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SEPTEMBER 1971

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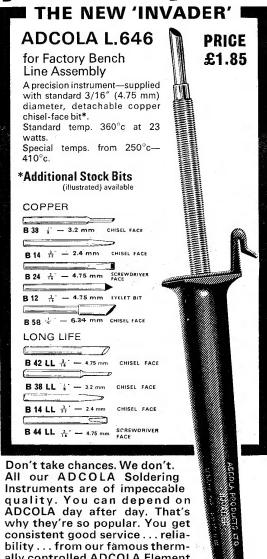
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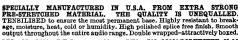


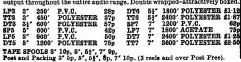
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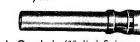
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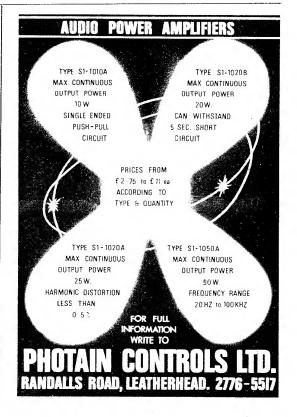
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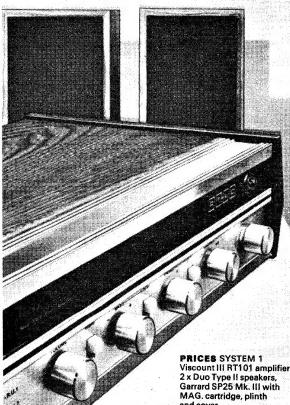
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and cover

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Total Available complete for only £52.00 + £2.50

### SPECIFICATION

SPECIFICATION

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16 10W 16 14Hz 0-19%. P.U.1 (for ceramic cartridges)
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Total £52.00

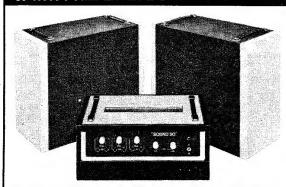
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to set up and can be relied upon to come over with all the quality and power you need.

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SPEAKERS! Size 20" × 20" × 10" incorporating 12" heavy duty 25 watt high flux, quality loudspeaker with cast frame. Cabinets attractively finished in two tone colour scheme—Black and grey.

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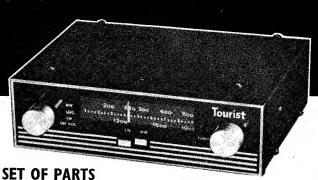
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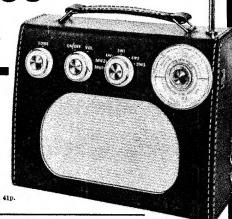
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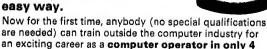
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ı	5 amp. A.C.* £2.60
Į	10 amp. A.C. #£2-60
ł	20 amp. A.C.*£2.60
1	30 amp. A.C.*£2-60
	bo amp. H.o. au ou
_	

1 amp. ....

# Type MR.52P. 22in, square fronts.

\$\frac{1}{2}\text{in. square fronts.}\$

10V. D.C. \$2.00
20V. D.C. \$2.00
50V. D.C. \$2.00
50V. D.C. \$2.00
15V. A.C. \$2.00
15V. A.C. \$2.00
SMeter ImA \$2.10
VU Meter \$3.10
1 amp. A.C. \$2.00
10 amp. A.C. \$2.00
30 amp. A.C. \$2.00
30 amp. A.C. \$2.00 

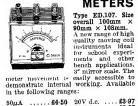
# Type MR.65P. $3\frac{3}{8}$ in, $\times$ $3\frac{1}{8}$ in, fronts.

$50\mu A$ $3.37\frac{1}{2}$	10V. D.C #2-10
50-0-50μA £2.75	20V. D.C £2·10
100μA £2.75	50V. D.C £2·10
100-0-100µA£2.60	150V.D.C £2·10
200µA £2-60	300V. D.C. \$2.10
500μA £2.37½	15V. A.C £2·10
500-0-500μA £2 10	50V. A.C £2·10
1mA £2·10	150V. A.C. £2·10
5mA £2·10	300V. A.C. £2·10
10mA £2·10	500V. A.C. £2·10
50mA £2:10	S Meter 1mA £2.371
100mA £2:10	VU Meter £3.371
500mA £2·10	50mA A.C. *£2.10
1 amp £2-10	100mA A.C.*£2·10
5 amp £2·10	200mA A.C.*£2:10
10 amp £2·10	500mA A.C.*£2·10
15 amp £2·10	1 amp. A.C.* £2-10
20 amp £2·10	5 amp. A.C.* £2.10
30 amp £2·10	10 amp. A.C.*£2-10
50 amp £2.37½	20 amp. A.C.*£2.10
5V. D.C £2.10	30 amp. A.C. *£2.10
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1A d.c £3.97	500mA/5A d.c.£4
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### Type MR.38P. 1 21/32in. square fronts 200mA . . . £1-37 300mA . . . £1-37 **₹** 1 amp. 2 amp. 5 amp. 1017 1017 1017 1017

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10 amp. 21.372
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100μA £1.871
100μA £1.871
100μA £1.872
200μA £1.75
200μA £1.75
200μA £1.75
200μA £1.50
100ν. D.C. £1.372
500μ-500μA£1.73
100ν. D.C. £1.372
100μA £1.50
100ν. D.C. £1.372
100μA £1.51
100ν. D.C. £1.372
100μA £1.371
100ν. D.C. £1.372
100μA £1.371
100ν. D.C. £1.372
100μA £1.371
100ν. A.C. £1.372
100μA £1.371
100μA £1.371
100μA £1.371
100μA £1.371
100μA £1.372
100μA £1.372
100μA £1.373
10μA £1.373
10μ

Type MR.45P. 2i	n. square fronts.
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50-0-50µA £2·10	10V. D.C £1.50
100µA £2·10	20V. D.C £1.50
100-0-100µA £1.871	50V. D.C £1.50
200uA £1.87	300V. D.C. £1.50
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500-0-500µA £1.50	300V. A.C. £1.50
1mA £1.50	S Meter 1mA £1.87
5mA £1.50	VU Meter £2.25
10mA £1.50	1 amp. A.C. * £1 50
50mA £1.50	5 amp. A.C.* £1.50
100mA £1.50	10 amp. A.C. *£1.50
500mA £1.50	20 amp. A.C. *£1.50
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2G309 30p 2N3566 22p 2S303 2G371 15p 2N3568 25p 2S304	57p BC125 15p BFX44 72p BC126 25p BFX68	87p NKT240 27p 87p NKT241 27p	CA3012 88p FJH171 25p SN7448 100p CA3013 105p FJH181 25p SN7450 25p	IS5 25p 30C18 75p EM87 43p IT4 25p 30F5 85p EV51 40p
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2N388A 49p 2N3572 97p 28503 2N404 21p 2N3605 27p 3N83 2N696 15p 2N3606 27p 3N128	27p BC136 15p BFX86 40p BC137 15p BFX87 70p BC138 37p BFX88	25p NKT244 17p 27p NKT245 20p 25p NKT261 20p	CA3018A FJH241 25p SN7454 25p 110p FJH251 25p SN7460 25p CA3019 84p FJJ101 50p SN7472 40p	2D21 35p 30FL14 75p EZ40 45p 3Q4 40p 30L15 85p EZ41 45p
2N697 16p 2N3607 22p 3N140 2N698 25p 2N3638 18p 3N141	77p BC140 35p BFX89 72p BC141 35p BFX93A	62p NKT262 30p 70p NKT264 20n	CA3020 126p FJJ111 50p SN7473 45p CA3020A FJJ121 60p SN7474 45p	384 35p 30L17 80p EZ80 25p 3V4 45p 30P12 80p EZ81 28p 5R4 60p 30P19 80p GZ32 48p
2N699 42p 2N3638A 20p 3N142 2N706 9p 2N3641 18p 3N143 2N706A 11p 2N3642 18p 3N152	55p BC147 15p BFY11 67p BC148 11p BFY18 87p BC149 15p BFY19	42p NKT271 20p 25p NKT262 20p	160p FJJ131 60p SN7475 100p CA3021 156p FJJ141 125p SN7476 45p	5U4 38p 30PL1 70p GZ34 60p 5V4 42p 30PL13 93p KT66 \$2.05
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2N718 25p 2N3645 25p 40251 2N718A 30n 2N3691 15p 40309	32p BC154 35p BFY29 32p BC157 15p BFY30	40p NKT281 27p 40p NKT401 87p	CA3028A 74p FJJ251 125p SN7492 100p CA3028B FJL101 125p SN7493 100p	6AC7 25p 35Z4 30p PC86 60p 6AG7 40p 35Z5 40p PC88 60p
2N726 25p 2N3692 18p 40310 2N727 25p 2N3693 15p 40311 2N914 17p 2N3694 18p 40312	45p BC158 15p BFY41 35p BC159 15p BFY43 47p BC160 35p BFY50	50p NKT402 90p 62p MKT403 75p 22p NKT404 62p	105p FJY101 25p SN7495 100p CA3029 87p IC10 250p SN7496 100p CA3029A IC12 250p SN74107 45p	6AK5 30p 50B5 45p PC97 45p 6AK6 57p 50C5 40p PC900 48p
2N916 17p 2N3702 12p 40314 2N918 30p 2N3703 10p 40315	37p BC167 15p BFY51 37p BC168B 14p BFY52	20p NKT405 75p 22p NKT406 62p	165p L900 40p SN74153 CA3030 137p L914 40p 190p	6AL5 20p 80 50p PCC84 40p 6AM6 33p 85A2 45p PCC85 40p 6AQ5 35p 807 50p PCC88 55p
2N929 22p 2N3704 12p 40316 2N930 24p 2N3705 10p 40317 2N937 52p 2N3706 10p 40319	47p BC168C 15p BFY53	17p NKT451 62p 57p NKT452 62p	CA3035 122p L923 40p SN74154 CA3036 72p MC724P 66p 220p CA3039 82p MC780P 247p SN74160	6AS6 37p 1625 50p PCC89 50p 6AT6 30p 5763 70p PCC189 55p
2N1090 22p 2N3707 12p 40320 2N1091 22p 2N3708 8p 40323	47p BC170 12p BFY77 32p BC171 15p BFY90	42p NKT453 47p 57p NKT713 20p 65p NKT717 42p	CA3041 109p MC788P 82p 180p CA3042 109p MC790P 124p SN74161	6AU6 25p 6146 £1.50 PCF80 30p 6AV6 30p AZ31 50p PCF82 34p 6BA6 25p CY31 35p PCF84 50p
2N1131 25p 2N3709 10p 40324 2N1132 25p 2N3710 10p 40326 2N1302 17p 2N3711 10p 40329	47p BC172 15p B8X19 37p BC175 22p B8X20 30p BC177 22p B8X21	17p NKT734 27p 17p NKT736 35p	CA3043 137p MC792P 66p 180p CA3044 120p MC799P 66p SN74164 CA3045 122p MC1303L 220p	6BE6 30p DAF91 25p PCF86 60p 6BH6 45p DAF96 42p PCF800 80p
2N1303 17p 2N3713 187p 40344 2N1304 22p 2N3714 200p 40347	27p BC178 20p BSX26 57p BC179 22p BSX27	37p NKT773 25p 45p NKT781 30p 47p OC16 50p	CA3045 122p MC1303L 220p CA3046 81p 262p SN74165 CA3047 137p MC1304P 225p	6BJ6 45p DF91 25p PCF801 50p 6BQ7A 40p DF96 42p PCF802 50p 6BR7 85p DK91 35p PCF805 75p
2N1305 22p 2N3715 222p 40348 2N1306 24p 2N3716 290p 40360	52p BC182 12p BSX28 42p BC182L 10p BSX60	32p OC19 37p 82p OC20 75p	CA3048 204p 275p SN74192 CA3049 160p MC1305P 225p	6BR3 65p DK92 50p PCF806 70p 6BW6 85p DK96 42p PCF808 75p
2N1307 24p 2N3773 240p 40361 2N1308 29p 2N3791 275p 40362 2N1309 24p 2N3819 34p 40370	47p BC183 10p BSX61 57p BC183L 10p BSX76 32p BC184 15p BSX77	62p OC22 50p 15p OC23 60p 20p OC24 60p	CA3050 185p CA3051 184p CA3052 165p	6BW7 70p DL92 35p PCL82 35p 6BZ6 35p DL94 45p PCL83 65p 6C4 33p DL96 42p PCL84 45p
2N1507 17p 2N3820 57p 40406 2N1613 21p 2N3823 75p 40407	57p BC184L 12p BSX78 40p BC186 25p BSY24	25p OC25 87p 15p OC26 25p	CA3053 46p MC1435P 162p CA3054 109p 845p TAA242	6C4 33p DL96 42p PCL84 45p 6CD6 \$1:15 DM70 32p PCL85 40p 6CL6 50p DY86 33p PCL86 45p
2N1631	52p BC187 27p BSY25 55p BC212L 12p BSY26 62p BC213L 12p BSY27	15p OC28 62p 17p OC29 62p 17p OC35 50p	CA3055 240p MC1552G 425p CA3059 165p 461p TAA243 150p CA3064 120p MC1709CG TAA263 75p	6CW4 63p DY87 85p PFL200 70p 6F1 62p E88CC 65p PL36 55p
2N1638 27p 2N3855A 30p 40412 2N1639 27p 2N3856 30p 40467A	50p BC214L 15p BSY28 57p BCY10 27p BSY29	17p OC36 62p 17p OC41 22p	FCH101 85p 94p TAA293 97p FCH111 105p MFC4000P TAA300 175p	6F6G 30p E180F 95p PL81 50p 6F13 38p EABC80 35p PL82 45p 6F14 65p EAF42 35p PL83 45p
2N1701 110p 2N3856A 35p 40468A 2N1711 24p 2N3858 25p 40528 2N1889 32p 2N3858A 30p 40600	72p BCY31 30p BSY36	25p OC42 25p 25p OC44 17p 25p OC45 12p	FCH121 105p 112p TAA310 125p FCH131 50p PA222 437p TAA320 72p FCH141 105p PA230 100p TAA350 175p	6F15 65p EB91 20p PL84 40p 6F18 45p EBC41 55p PL500 75p
2N1893 37p 2N3859 27p 40603 2N2147 72p 2N3859A 32p AC107	50p BCY33 20p BSY38 30p BCY34 25p BSY39	20p OC46 15p 22p OC70 15p	FCH151 105p PA234 100p TAA435 147p	6F23 80p EBC81 30p PL504 80p 6H6 20p EBF80 40p PY32 55p 6J4 50p EBF83 40p PY33 63p
2N2160 57p 2N3860 30p AC126 2N2193 40p 2N3866 150p AC127 2N2193A 42p 2N3877 40p AC128	24p BCY39 60p BSY51	50p OC71 12p 32p OC72 12p	FCH171 105p PA246 245p TAA522 380p FCH181 105p PA424 285p TAA530 495p	6J5 20p EBF89 32p PY80 35p 6J5GT 30p EBL21 60p PY81 30p
2N2194 27p 2N3877A 40p AC151 2N2194A 30p 2N3900 37p AC152	18p BCY41 15p BSY53 22p BCY42 15p BSY54	82p OC73 30p 87p OC74 30p 40p OC75 22p	FCH191 105p PA264 447p TAA811 445p FCH201 130p PA265 497p TAB101 97p FCH211 130p SN7400 25p TAD100 150p	6J6 20p EC86 60p PY82 30p 6J7 45p EC88 60p PY83 38p 6K8G 35p ECC40 60p PY88 40p
2N2217 27p 2N3900A 40p AC154 2N2218 29p 2N3901 97p AC176 2N2219 31p 2N3903 25p AC187	22p BCY43 15p BSY56 22p BCY54 32p BSY79	90p OC76 22p 45p OC77 30p	FCH221 130p SN7401 25p TAD110 197p FCH231 150p SN7402 25p SL403A 187p FCJ101 160p SN7403 25p SL702C 147p	6L6GT 45p ECC84 80p PY800 50p 6LD20 40p ECC85 60p PY801 50p
2N2220 25p 2N3904 25p AC188 2N2221 25p 2N3905 30p ACY17	27p BCY59 22p BSY95A 27p BCY60 97p C424	57p OC78 20p 12p OC81 20p 15p OC81D 20p	FCJ111 150p SN7404 25p UA702A 280p FCJ121 275p SN7405 25p UA702C 77p	6Q7 40p ECC88 40p U25 75p 6SA7 40p ECF80 35p U26 75p 6SG7 35p ECF82 35p U50 32p
2N2222 29p 2N3906 30p ACY18 2N2222A 35p 2N4058 16p ACY19	24p BCY70 20p C450 24p BCY71 30p GET102	15p OC82 25p 30p OC82D 15p	FCJ131 275p SN7406 80p UA703C 137p FCJ141 525p SN7408 25p UA709C 125p	6SJ7 40p ECF86 65p U52 33p 6SK7 35p ECH21 57p U191 75p
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# PRACTICAL

VOL 47 NO 5

Issue 775

SEPTEMBER 1971

# TOPIC OF THE MONTH

# **Electronic Vandals**

NEW generation of vandals is infesting the country-side. The scale of the menace can be judged by a report in The Times which says that more than 50 archaeological sites in Britain have been robbed by vandals using metal locators in the "last three months". These looters are urged on by irresponsible promptings such as contained in the book A Fortune Under Your Feet, which would not have been publicised in the article last month had the editor been aware of passages which recommend amateurs to operate metal locators on established Roman and prehistoric sites. We also condemn remarks made recently in Coin Monthly which, in slightly hysterical rhetoric, suggest that "an insignificant group of professional archaeologists" want to stop "those rights and pleasures" (of operating metal detectors). This technique of twisting the facts is as familiar as it is sickening.

This type of rabble-rousing emotionalism, coupled with the stimulation of the greed motive, is a ready incitement to gullible and unprincipled people to abandon whatever vestiges of conscience they may have had. It has led to a situation where our dwindling number of archeological sites are being systematically ravaged. Archaeologists are seriously considering the possibility of not releasing information of newly discovered sites-a sad reflection of the times we live in, since the main objective of such explora-

tion is the dissemination of knowledge.

It is galling also that having developed the skill of scientific and methodical excavation to its modern level, yielding a degree of information hitherto impossible, a process made possible largely by generations of voluntary research, skilled field workers are now confronted by sites irrevocably ruined by zombies whose motivating objectives are simply personal gain. A sad, sordid picture.

There are laws to protect important sites and moves are afoot for heavily increased penalties for despoiling archaeological sites. Welcome as this might be, we feel that the problem is more a case of civil conscience than Acts of Parliament. Readers of P.W. can help. Avoid areas where invaluable evidence may be disturbed or destroyed. If a "casual" find looks important, contact your county archaeological society at once—and don't spread the word around; if you have discovered a site of potential importance it will require highly specialised workers and scientific treatment to interpret the findings and to extract the maximum information. This may be impossible for all time if a site is heedlessly disturbed. Moreover, if you are genuinely interested in the advancement of knowledge, coupled with the excitement of glimpses into the past, -continued on page 418

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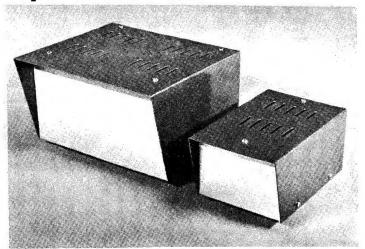
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OCTOBER ISSUE WILL BE **PUBLISHED ON SEPTEMBER 10th** 

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# NEWS... NEWS... NEWS...

# Impex instrument cases



The cases shown come in a standard range of 36 sizes and are of 20 s.w.g. steel with a stove enamelled green hammer finish. The passivated zinc-plated steel base is 18 s.w.g.

Detachable front and back panels are of 18 s.w.g. satin anodised aluminium and are designed to lie flat for easy piercing. Phillips screws are used throughout and each case is provided with four rubber feet.

At no additional cost, Impex cases can be supplied with all or some of these specifications: alternative colours of blue or bronze; no hood; no ventilating slots or back/front anodised front/rear panels.

Examples of prices are:  $2^{1}_{2}$ in. height,  $3^{3}_{8}$ in. width and depth of 5in.: £1·13 plus 15p post and packing. A case measuring 5in× $13^{1}_{2}$ in.×5in. costs £3·19 plus 25p p.&p. and one measuring 5in.× $13^{1}_{2}$ in.×11in. costs £6·40 with p.&p. charge of 28p.

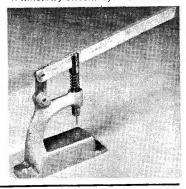
Further details of prices of these cases, one-off jobs and rack-mounting units may be obtained from "Impex" McArdle and Brainsby (Import & Export) Ltd., P.O. Box 2BB, Newcastle-upon-Tyne, NE99 2BB.

# **Precision hand press**

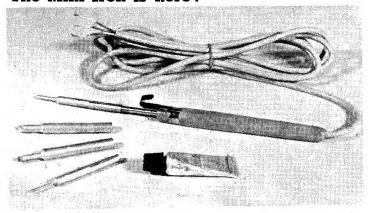
The "Innovcom" precision hand press can be used for the assembly and 'dis-assembly of small components and the application of precise pressure to a very small surface area.

Construction is light alloy body with hollow open base, drilled for two screw attachment to bench and drilled beneath chuck to hollow cavity. Link and plunger are of mild steel. Sintered bronze Oilite bushes are oiled for life. Handle length is 7½in. and full handle traverse of 2½in. gives chuck movement of 0.6in. Daylight (clearance) from chuck to platform is lin. Chuck takes 40 thou wire, nail shaft or rod.

The price is £3.45 postage and packing included and the unit is for UK distribution only. Innovcom Ltd., Southbank, Daveylands, Wilmslow, Cheshire, SK9 2AG.



# The Mini Iron is here!



The Adamin Model 15 is a miniature soldering iron which comes complete with a set of interchangeable bits and a tube of lubricant. This iron weighs half an ounce less flex but its versatility is by no means "light weight." Used with the interchangeable bits from 0 047in. to  $\frac{2}{16}$  in. it is suitable for all jobs from minute hearing aid soldering to heavy "chassis bashing."

A 12V version of the iron is also available and the complete hobby pack as shown and priced at £2·30 (p.&p. free) is available from Light Soldering Developments at 28 Sydenham Road, Croydon CR9 2LL.

# NEWS... NEWS... NEWS...

# Transensor kit

In the February 1971 issue we mentioned the Transensor slide synchroniser, which enables a cassette recorder to be used with an automatic slide projector.

The unit is now available in kit form, enabling it to be fitted by the purchaser to his own cassette tape recorder. The makers claim that some twenty or thirty improved versions of the mono cassette recorder appear on the market every year and as the Transensor S/2 kit is adaptable, it can be transferred from one machine to another as new models are introduced. The price is £18.

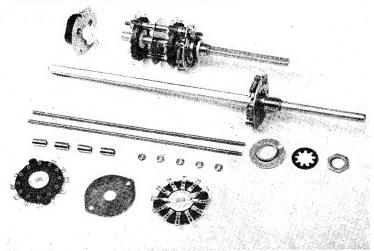
One other feature of the Transensor S/2 is that it allows pulses to be introduced and erased independently of the master sound track.

Further information may be obtained by writing to Audio Visual Picture Enterprises, 17 Abercorn Place, London, N.W.8. or telephoning 01-328-1461.

# Bib auto record cleaner

The Bib Division of Multicore Solders Ltd., Hemel Hempstead, Herts has introduced an automatic record cleaner under the brand name Bib Groov-Kleen Model 40. Resembling a high quality miniature cartridge arm, it is finished mainly in anodised aluminium. The base is supplied with a self-adhesive disc so that it is easily fixed to the player deck. The base has a chromiumplated pivot pillar which can be raised or lowered so that the arm can be adjusted to be parallel with the turntable. The arm, which is cranked to provide better tracking, has a brush at one end and a counterweight at the other. A small roller mounted behind

# Return of the Maka Switch



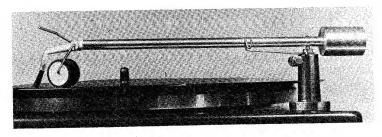
Constructors who have mourned the loss of the MAKA switch will be delighted to learn that A. B. Electronic Components Limited manufacture a similar, but more versatile switch, which is available from stockists, in kit form. The size is approximately the same as the MAKA, and as can be seen from the illustration consists of a shafting unit with a six inch length shaft, wafers, screens, spacers, studding and a mains switch.

The six inch shaft length will accommodate several wafers, or it may be cut down if only one or two wafers are required. The wafers are available as follows:—"Break before Make"—one pole twelve way; two pole six way; three pole four way; four pole three way; six pole two way. "Make before Break"—one pole ten way; two pole five

The metal screens are a feature which have been added. These switch kits are available from these suppliers:—Crescent Radio, 40 Mayes Road, London, N.22; Home Radio (Components) Ltd., 240 London Road, Mitcham, Surrey; Servio Radio, 156-158 Merton Road, Wimbledon, London, S.W.19; Garland Bros., Chesham House, Deptford Broadway, London, S.E.8; Radioparts, 5 Market Way, Plymouth, Devon.

the brush automatically sets its own level. A swivelling arm-rest is provided to hold the arm when a record is being placed on the turntable.

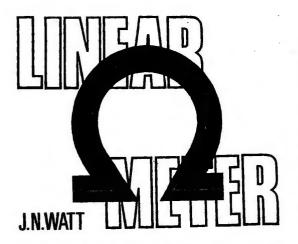
The recommended Retail Price is £2.08 plus P.T. of 51p. Bib Division, Multicore Solders Ltd., Hemel Hempstead, Herts.



# **Derby Rally**

The fourteenth annual Derby Mobile Radio Rally will be held on Sunday August 15th, 1971, at Rykneld School, Bedford St., Derby. There will be a band concert, junk sale, prize draw, children's events, trade stands, competitions, demonstrations, ice cream, refreshments, etc. in addition to G3ERD/A on 160 metres and G8DBY on v.h.f.

Admission and parking are free and further details may be obtained from Tom Darn G3FGY, Hon. Organiser, "Sandham Lodge," 1 Sandham Lane, Ripley, Derbys. DE5-3HE. Phone Ripley 2972.



THE first item of test equipment obtained by most home experimenters is a multi-range testmeter, capable of reading voltage and, probably, current and resistance as well. One large drawback of the ohms ranges is that they are very non-linear, being open at the low end and extremely cramped at the high end, see Fig. 1.

Besides being inconvenient to read, accuracy suffers when measuring values of resistance more than, say,  $100 \mathrm{k}\Omega$  on most instruments. Moreover, increasing resistance is read at the left-hand end, the opposite to the voltage and current ranges.

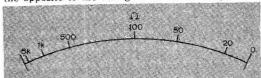


Fig. 1. Illustrating the cramped scale of a normal ohm-meter.

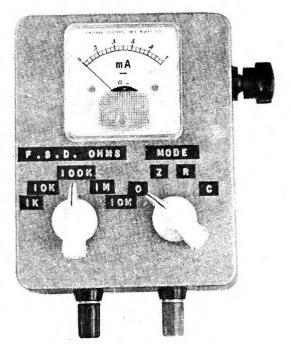
The ohm-meter described here has a linear calibration, with increasing resistance towards the righthand end of the scale and uses five switched ranges to cover  $10 \mathrm{M}\Omega$ ; the lowest value of resistance that can be measured is set by the accuracy with which the meter can be read but is about  $100\Omega$ . The possibility of extending this lower end downwards by a factor of 10 is discussed later. The ohm-meter is self-contained, although external batteries and an external moving-coil meter could be employed.

As a bonus, the form of circuitry used lends itself to adapting the instrument to test capacitors as well. Their value can be read and an indication is given of their insulation resistance i.e. leakage. The range of capacitors that can be tested in this way is about  $0.1\mu F$  to  $10,000\mu F$ , including electrolytics, but excluding those of less than about 8 volt rating, unless precautions are taken (see later for further details on this point.)

# PRINCIPLE OF OPERATION

The principle of operation of the linear scale ohmmeter is illustrated in Fig. 2. A constant current is sent through the unknown resistance,  $R_{\rm X}$ . Now, by Ohm's Law:—

$$R_x = \frac{E}{I}$$



so that if I is constant:-

$$R_x \propto E$$

and a measure of E is a measure of  $R_{\rm X}$ . Thus if a linear scale voltmeter can be arranged to read  $R_{\rm X}$ , this also will be read on a linear scale.

To measure capacitance, use is made of the relationship

$$Q = VC$$

Charge Q is current x time, so

$$VC = It$$
or  $C = \frac{It}{\nabla}$ 

Remembering that C is in farads and I is in amperes, a current of  $1\mu A$  flowing into a capacitor C for 10 seconds and producing a rise in potential of 10 volts means that C has a value:—

$$C = \frac{10^{-6} \times 10}{10}$$
$$= 1\mu F$$

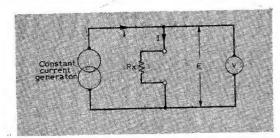


Fig. 2. Circuit to demonstrate the principle of the linear scale ohmmeter.

Similarly,  $10\mu A$  for 10 seconds and a rise of 10 volts means that a capacitor of  $10\mu F$  was connected, and so on.

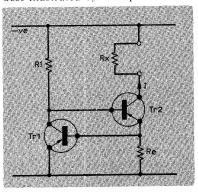
Actual values of current used and voltage monitored are different from these examples but the principle is exactly the same.

Two difficulties now present themselves. How can we generate a constant current? How can we measure E on the voltmeter without upsetting the circuit? Remember that simple voltmeters, based on a moving coil meter and series resistors must draw some current to function at all. This latter point is best illustrated by example.

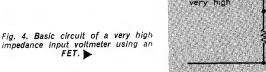
if Tr2 has a high enough current gain for base current to be ignored, a constant current will flow at Tr2 collector. This will remain so, no matter if the value of  $R_{\rm X}$  varies, due to the negative feed-back loop between Tr2 base and Tr1 collector, which functions as follows.

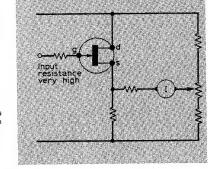
Suppose, for any reason, I increases. Then the voltage developed across  $R_{\text{o}}$  will increase, and so therefore will the base current of Tr1. The resultant voltage drop at Tr1 collector will tend to reduce the base current of Tr2 and hence also reduce its emitter current, so counter-acting the original increase.

The actual value of constant current is set to



▼ Fig. 3. Circuit of a constant-current generator.





Suppose  $R_X$  is  $1M\Omega$  and we have a  $100\mu A$  meter, which requires  $100k\Omega$  in series to read as a 10 volt f.s.d. voltmeter; if this were placed in parallel with  $R_X$ , as in Fig. 2 then obviously it would shunt the  $1M\Omega$  to a considerable extent, and give a false reading.

To deal with the point first raised, that of generating a constant current, consider the circuit of Fig. 3. The base-emitter voltage of Tr1 is virtually constant at about 0.7 volt for a silicon transistor; hence the voltage drop across  $R_{\rm e}$  is similarly constant. With a constant voltage across  $R_{\rm e}$ , a constant current must flow to the emitter of Tr2 and hence,

that required by adjusting the value of  $R_e$ , and also by making some broad adjustment to R1. A value for R1 suitable for  $1\mu A$  is no longer so when, say, 10mA is required, and so some changes in R1 are called for.

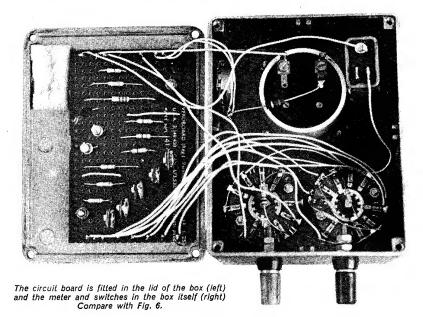
While any desired current can, in theory, be derived, obviously there are practical limitations. At the upper end dissipation in the transistors will be the deciding factor; note that maximum transistor dissipation occurs when  $\mathbf{R}_{\mathbf{X}}$  is at its lowest value, for then, for a given (constant) current, the voltage across the transistor Tr2 is greatest.

The lowest current that can be set up is limited

in two ways. Adequate current gain must be available and the desired current must be large compared to the leakage current.

In the practical circuit, a lowest current of about  $0.7\mu$ A is used; a BCY71 has a current gain of about 40 at this level, while leakage current is quoted as 0.05 pA maximum. Should a substitute transistor be used, constructors must ensure that it is suitable from these points of view. No germanium device is likely to be a practicable substitute; BC179, BCY70 and BC157 should prove to be reasonable alternatives.

With a constant current flowing through  $R_x$  we now require a high impedance voltmeter to measure the voltage



drop across  $R_x$ , so giving its resistance value. An FET seems to be called for and the ready availability of these devices at reasonable prices makes one the obvious choice.

Many circuits of high impedance voltmeters using FETs have appeared in the technical press but the one used here is based on a simple configuration published by Mullard Ltd., see Fig. 4.

The FET is arranged as a source follower; its source voltage is compared with a voltage derived from a potential divider. The series resistor in the gate circuit gives some protection against unwanted transients, which could damage the gate-source junction, by limiting the gate current. The input resistance of the circuit of Fig. 4 is many tens of megohms; in point of fact, the author was unable to measure it, it was so high—just what we require.

# PRACTICAL CIRCUIT

The full circuit diagram now becomes as in Fig. 5, which includes the necessary switching.

One pole of S1 selects different emitter resistors for Tr2, consisting of a fixed and small variable in series, and this is the range switching. The other pole of S1 switches in one of three different resistors, to cover the five decades, for Tr1 collector load.

With S2 in the "Z" position, R9 is connected from the FET gate to supply negative, so permitting zero setting of the meter to be done by means of VR6. This done, and with the test resistor connected, when S2 is moved to position "R", the value of the resistor is displayed on a linear scale. This means that, for example, on the  $10 \mathrm{k}\Omega$  range, a reading of  $0.7 \mathrm{m}\mathrm{A}$  is to be taken as indicating  $7 \mathrm{k}\Omega$ , a reading of  $0.3 \mathrm{5m}\mathrm{A}$ ,  $3.5 \mathrm{k}\Omega$  and so on.

# 

# \* components list

R1	68 Ω	R6	6-8kΩ	R11 47kΩ
R2	680 Ω	R7	68kΩ	R12 2.7kΩ
	6-8kΩ		680k Ω	R13 680 Ω
R4	68k Ω		1kΩ	R14 2.2kΩ
R5	680k 42	R10	68k Ω	100000000000000000000000000000000000000
All !	W 10%	-10	30.00	300 Co. 100 Co.
VRI	100 Ω mini	ature j	ore-set	
	1\Ω		14	The second second
	10kΩ ,,		4.6	Service Control of
	100kΩ .,		41	
	1MΩ		6	
VR6	4.7k $\Omega$ (or	5k (1) (	carbon poi	tentiometer
Semi	conductor			
Tr 1	BCY71	Tr 2	BCY71	Tr 3 2N3819
	llaneous:			TENEDO DO
wafe	er switch. N	11, 0-1 r 2+in. ^	nA moving Veroboard	n. S2, 2 pole 4-wa g coil meter Dieca: 0-15in matrix. Te , PP3 battery, etc

For testing capacitors, the capacitor is connected with S2 at "Z", whence R9 removes any charge on the capacitor, and, after setting of zero as before, S2 rotated to "R" for either 10 seconds or until the meter reads f.s.d. Further rotation of S2 to "C" holds the reading and enables an assessment of capacitor quality to be made; see later under 'Capacitor evaluation'.

The second pole of S2 is the supply on/off switch. Note that S2b must be a make-before-break switch, otherwise the instrument will be switched off between

positions of S2 and this would invalidate any capacitor measurements. Should such a switch not be available, then S2b should be replaced by a small toggle or slide switch mounted conveniently; S2a can be the more usual break-beforemake type.

# CONSTRUCTION

Having dealt with the basic theory and circuit of the linear scale ohmmeter, we can turn to constructional details.

The photographs and drawing show the layout of the prototype, which was housed in a die-cast box.

Most of the components are mounted on a piece of 0.15'' pitch veroboard which is in turn mounted

Fig. 5. Complete circuit of the linear ohm-meter, with transistor lead-out connections.

Fig. 6. Suggested layout of components with the circuit board in the lid of the die-cast box.

on the inside of the box cover, using nylon nuts as insulated spacers. Connections to the switches, meter, battery and terminals are made via veropins. The small pre-set variable resistors, VR1 to VR5 are 0·1 watt components intended for 0·1" pitch veroboard, but, by a slight bending of the terminal tags can be made to fit the board used.

A suggested layout is given, Fig. 6, but if for any reason the constructor wishes to alter this in any way, then this is in order.

The dimensions of the box used will depend on a number of factors. The meter is the most important of these; in the prototype the meter

was one 2in. square. Although this may seem rather small, it is still capable of being read to about 2%, which is better than the accuracy with which the instrument can be set up by most constructors; it is also about the limit of accuracy of the meter movement itself.

A meter of such a size enables a box of  $4^3_4$  x  $3^3_4$  x  $2^1_4$ in. to be used, provided reasonable care is taken in the choice of components.

The rotary switches shown are of 1½in. diameter and so can be accommodated comfortably, while a PP3 9 volt battery is simply fitted by clamping it, in a vertical position, between the box and its cover. Two scraps of foam plastic, one at each end of the battery, prevent it from moving and at the same time provide insulation for its terminals. Connection to these is by soldering wires directly to them, there being no room for the usual snap connector. Since the whole instrument draws only 3 mA, plus any test current, a PP3 should last a very long time if the unit is switched off when not actually in use, and thus solder connections to the battery are no great disadvantage.

With wiring up completed, check that the voltmeter section is functioning correctly by switching S2 to "Z" and seeing that VR6 enables the meter to be brought to zero.

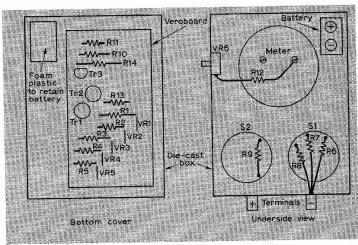
Actual setting up of the emitter resistors of Tr2 by means of VR1 to VR5 can only be done by making use of known values of resistance. However, even just the loan of resistors of 2% accuracy will suffice, while 5% resistors can be bought quite cheaply. Values required are  $10 \mathrm{M}\Omega$ ,  $1 \mathrm{M}\Omega$ ,  $100 \mathrm{k}\Omega$ ,  $10 \mathrm{k}\Omega$  and  $1 \mathrm{k}\Omega$ .

Start with range 1 and fit the  $1k\Omega$  to the terminals; with S2 at "R" adjust VR1 to give a reading of 1mA on the meter. Repeat on the other ranges, using the appropriate resistors, and remembering to put S2 to "Z" before removing resistors.

# CAPACITOR EVALUATION

Measurement of the value of medium to large capacitors and an indication of their leakage resistance can be considered as a bonus with the ohmmeter but can be omitted if desired.

Demonstration of this mode of use can best be carried out using a  $1\mu F$  paper capacitor. With S1 on range 5 and S2 at "Z", connect the capacitor and then put S2 to "R". Note that the meter reading



slowly increases at a constant rate. When the meter reads 0.8 mA or so, switch to "C"; the unit is now reading the voltmeter that was developed across the  $1 \mu F$  by the constant current which flowed into it on "R", and if the meter reading stays constant, or almost so, then the  $1 \mu F$  capacitor is a good one. The implication is that the voltmeter input resistance is very many megohms and hence is extremely difficult to measure.

With this confirmation that the  $1\mu F$  capacitor is a good, low leakage specimen, it is now an easy matter to measure its actual value.

First, put S2 back to "Z", so discharging the capacitor by means of R9. Then, switch S2 to "R" and commence timing, either by stop watch or a sweep second hand on a watch or clock. When 10 seconds have elapsed, or the meter reads f.s.d. which ever is the sooner, switch S2 to "C", noting the elapsed time if less than 10 seconds.

If, after 10 seconds, the meter reads 1mA, then the capacitor was exactly  $1\mu F.$  A meter reading of, say, 0.8 mA after 10 seconds indicates a value of

$$\frac{1}{0.8} \times 1$$
 or  $1.25 \mu F$  and so on.

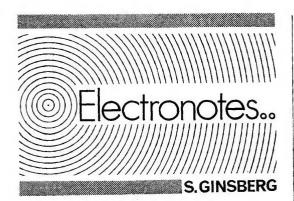
On the other hand, if 1mA was reached after only 7 seconds, then the capacitor was

$$\frac{7}{10} \times 1 \text{ or } 0.7 \mu\text{F}$$

Whether the capacitor is actually larger or smaller than the  $1\mu F$  corresponding to f.s.d. on range 5 is easily remembered by recalling that a large value capacitor will take longer to reach a given voltage at a given constant current than will a small value capacitor.

The values of capacitance corresponding to each position of S1 and 10 seconds for meter f.s.d. are as follows:—

RANGE	RESISTANCE	CAPACITANCE
1 2 3 4 5	1kΩ 10kΩ 100kΩ 1MΩ 10MΩ	$egin{array}{l} {f 10,000} \mu {f F} \ {f 1,000} \mu {f F} \ {f 100} \mu {f F} \ {f 10} \mu {f F} \ {f 1} \mu {f F} \end{array}$



HEN someone takes an accepted practice and does exactly the opposite the results are often astounding. This has been the case in the Siemens Laboratories in Berlin. Ever since the electron microscope has been invented, everyone has employed it to peer at the most minute items, relying on the microscope, with its tremendous magnification, to give a clear and detailed image. Siemens decided it was time to try things in reverse, rather like looking through the other end of a telescope. The result is that just as the items could be magnified by very large amounts they can also be reduced. This means that it is possible to put the whole of the contents of the Bible on an area of some 0.25×0.25mm. Think of the idea as a means of storage. Around 1,000 Biblelength volumes could be stored, ready to be read by looking at them through the electron microscope the "right way", in an area of 25 square millimeters. Scotland Yard might well be interested since it is possible to store one million photographs on a piece of foil 5×5mm. They'll be fingerprinting fleas

Valve enthusiasts, certainly those interested in generating r.f., have wagged their wise old heads for years at the transistor lovers and pointed out that when it comes to generating lots of power, the solid state devices available just aren't in the same class as their glass-encapsulated brothers. It now looks as though this is no longer true. One company has announced an h.f. amplifier which is broadband from 1.5-30MHz and thus requires no tuning. It is solid state-not a valve in sight-and will supply 1kW on all modes of radiotelegraph and radiotelephone transmissions. For the 1,000W output, which, by the way, is the continuous rating, it requires only 100mW input. The amplifier is not a delicate device. A patented level control will prevent any damage which might be caused by an antenna mismatch which can range from an open circuit to a short circuit.

Hats off to the Allen Company in America. They are making an electronic organ which uses a computer to provide tones and voices. The 22 circuits on chips half-an-inch square contain some 48,000 transistors. The various waveforms and tones are simply stored in the musical computer's memory and called out or "read" at will. The number of voices available are almost unlimited, running into millions. The electronics for the computer have been worked out in association with another American company-North American Rockwell who have done a great deal of work on aerospace digital computers.

It is essential to observe polarity when testing electrolytic capacitors.

Should these be of a very low voltage rating i.e. less than, say, 8 volts, then do not allow the meter reading to rise above about half scale. This will restrict the voltage on the capacitor to a low level. Using a charge time of 5 seconds in lieu of 10 seconds, then a meter reading of 0.4mA indicates a capacitance of

$$\frac{1}{0.4} \times 5 \text{ or } 1.25 \mu\text{F}$$

To sum up, the value of capacitance is, in general:

$$C_x = \frac{1}{I} \times \frac{t}{10} \times C$$

where I is meter reading in mA t is time to reach that reading

C is capacitance given in the table. A little practice with capacitors of known value will soon make this method clear.

It is always as well to carry out a quick check first on doubtful capacitors to test for leakage. If leakage is taking place, as indicated by a falling meter reading on "C", then any capacitance values read off will of course be in error, for some of the current fed to the capacitor will have gone towards providing the leakage current and the voltage reached will thereby be less.

Whether the amount of any leakage so detected can be tolerated will depend on the application intended and no general rules can be laid down.

# NOTES

If a multi-range current meter is available, it is quite instructive to use it to measure the test current on each range. No matter whether, say, on range 3,  $100k\Omega$  or a short circuit is connected, then about 60 µA should flow, and, similarly, with other currents on other ranges as appropriate. The actual value of the current depends on the particular setting of the appropriate emitter resistor.

În the introduction, mention was made of the possibility of extending the lowest range of resistance measurement downwards by a factor of 10. If carried out, this would make the lowest range indicate up to  $100\Omega$  with a good indication down to 10Ω. This is not possible, however, without some

alteration of transistor type.

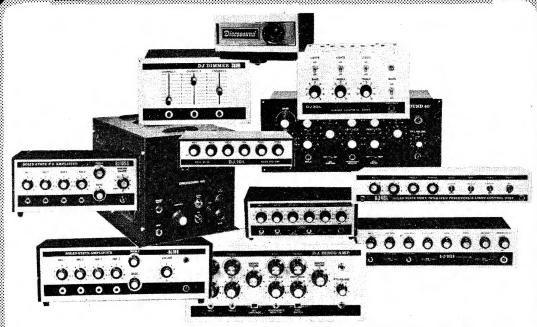
A range of 100Ω f.s.d. requires a constant current of about 60mA and so a possible dissipation in Tr2 of about 600mW, which is too great for the BCY71 specified.

A glance at some manufacturers' data shows that a number of suitable transistors are available, such as BFX88, 2N1132 and 2N4036 and no doubt others. Not all of these have a low enough leakage current however, to enable them to give good results on the  $10M\Omega$  range and it would be as well to omit this latter if the  $100\Omega$  range is incorporated.

The author regrets that he can give no further practical details, not having tried a low resistance range; what is obvious is that a larger battery, or a small mains supply, will be needed to provide the extra current required. The care that will be called for in use should not be overlooked, for up to 600mW might be dissipated in the test resistor, and consequently testing will have to be restricted to resistors of that wattage rating or greater.

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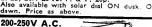
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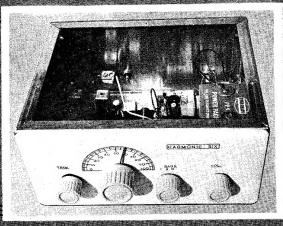
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# HARMONIE SIN DICHTIE EGRAVER GROGER



THIS is a short wave receiver covering the most active short wave bands of about 5·5-22MHz, or 55-14 metres, where very many short wave broadcasts are to be found. The range is divided into two bands, so related that harmonic mixing makes it unnecessary to have oscillator coil switching. This will be found to give excellent results with a minimum of constructional difficulty, and bandchange switch wiring is extremely simple.

The receiver is constructed in three sections, mixer, i.f. amplifier and audio amplifier, accommodated in a neat aluminium case which also takes the battery. A separate speaker is used, because headphones may sometimes be preferred, allowing personal short wave reception without disturbance to others.

# CIRCUIT DETAILS

**Mixer.** Fig. 1 is the circuit which also shows the division into three parts. The mixer assembly is complete with tuning capacitor, band-switch, coils and other items, having a small sub-panel for the controls.

L1 is the aerial coil for the higher frequency range of about 11-22MHz, while L2 is for about  $5\cdot5\text{-}11\text{MHz}$ . S1 selects the wanted coil. Base coupling windings for Tr1 are in series, and thus require no switching.

L3 is the oscillator coil, the values of C6 across the coil and C7 in series with the oscillator tuning capacitor VC2 allowing this circuit to tune from approximately 6MHz to 11·5MHz. When L2 is in

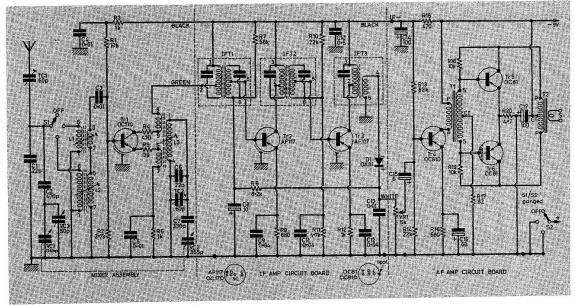
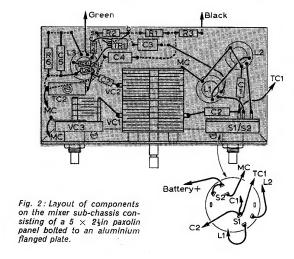


Fig. 1: Circuit diagram of the Harmonic Six with transistor lead-out connections.



circuit, VC1, in conjunction with C1 and C2, allows the aerial section to tune from approximately 5.5-11MHz. In terms of actual frequencies, the oscillator is 465kHz h.f. of the aerial circuit, in the usual way.

To reduce oscillator pulling, some transistor receivers have the oscillator working at half frequency, for the highest frequency range. This is purposely done in this receiver, so that no switching, or further coil or trimmer will be needed instead of L3. Instead, the second harmonic of the oscillator frequency is used, and the range of this harmonic is approximately 12-23MHz. L1 is in circuit for this band, and tunes 465kHz lower in frequency than the second harmonic, or about 11·5-22·5MHz. There is

actually a little overlap between bands, giving continuous coverage from about  $5\cdot 5\text{-}22MHz$ .

The receiver can bring in many transmissions with a short indoor aerial, either wire, or self-supporting rod, and the aerial trimmer VC3 is a panel control, so that L1 or L2 can be peaked for maximum results, despite changes to the aerial.

Resistors R4 and R5 are to avoid excess oscillation, which sometimes causes whistles and other troubles on the higher frequencies. In some cases it may be worth while modifying the values of R4 or R5, as mentioned later.

The mixer assembly is secured in the case by the bushes of VC3 and S1, this completing the positive or earth return. Two leads run from the mixer assembly—black for the negative supply circuit, and green for the i.f. amplifier.

IF Circuit. This board carries i.f.t.1, i.f.t.2, i.f.t.3, Tr2, Tr3, and the associated components. It is wired separately, and may be checked or tested alone. Two double-tuned i.f.t.'s with AF117's give very good gain and selectivity. Automatic gain control bias is obtained from the diode D1 in the usual way.

The positive circuit of the i.f. board is completed by its two mounting bolts, through the metal case. Audio output is from D1, the white lead running to the volume control VR1.

VR1 is fixed to the case front and so is not present on either i.f. or a.f. boards.

**AF** Amplifier. Audio signals are taken to the audio amplifier and driver Tr4, which is followed by the push-pull output pair Tr5/6. This is a straightforward arrangement capable of giving a good output, and it is wired complete on its own circuit board. Positive returns are made through the fixing bolts and metal case, as before.

S2 is the second pole of the switch S1, which is 3-way. The third position interrupts the positive

circuit for "OFF," Should an on-off switch be preferred on VR1, it is necessary to fit a compact potentiometer with switch, or there will not be enough space for a PP9 battery in the receiver.

A connecting point is provided at the junction of C14 and R15 for the black lead from the i.f. board.

# the i.f. board. CONSTRUCTION

Mixer Assembly. Components are placed as in Fig. 2. Holes are made in the 5 x 2in. flanged plate for the variable capacitors and switch. Washers or other means of spacing, about <sup>3</sup>8in. thick, are put between the plate and ganged capacitor, on each of the three fixing screws. It is essential that these screws are not too long.

Two bolts hold the insulated board to the flanged plate. Wire

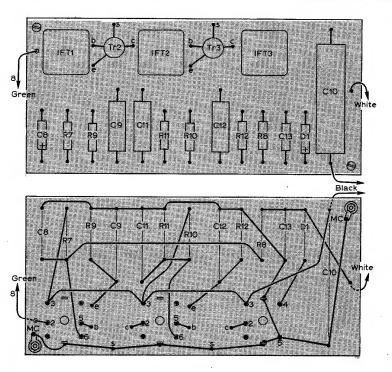


Fig. 3: Component layout and wiring details of the i.f. circuit board  $3\frac{3}{4} \times 1\frac{3}{4}$  in.

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AC176 25p BF170 35p OC200 40p AC187 30p BF173 30p OC201 70p AC188 30p BF177 40p OC202 80p	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	25 + 25p 100 + 23p 500 + 21p 1000 + 18p	25 + 12p 100 + 10p 500 + 8p
ACY17 30p BF178 25p OC203 40p ACY18 25p BF179 40p OC204 40p ACY19 25p BF180 35p OC205 75p ACY20 20p BF181 35p OC206 90p	ZENER DIODES 1.5 AMP MINIATURE WIRE ENDED PLASTIC 400 M/W 5% Miniature	0C45 Mullard 15p	ORP12 Mullard 50p 25 + 45p
ACY21 20p BF182 30p OC207 90p ACY22 10p BF184 20p OCP71 97p	Type P.I.V. 1-49 50+ 100+ 500+1000+ Volt. 15p each. PL4001 50 10p 9p 8p 7p 6p 25+ 12p; 100+ 10p; 500+ 8p;	25 + 13p 100 + 12p	25 + 45p 100 + 42p 500 + 40p
ACY39 50p BF185 20p ORP12 50p ACY40 15p BF194 17p ORP60 40p AD140 50p BF195 15p ORP61 42p AD149 50p BF196 15p TIP29A 50p	PL4003 200 12p 11p 10p 9p 8p 1½ Watt 5% Metal case. Wire Ends. PL4004 400 12p 11p 10p 9p 8p 6.8V. all voltages to 100V. 25p each.	500 + 10p 1000 + 8p	2N930 25p
AD161 37p BF197 15p TIP30A 60p AD162 37p BF200 37p TIP31A 62p AF114 25p BF274 37p TIP32A 75p	PL4005 800 1ab 13b 11b 10b 9b 12b 12b 12b 10b 12b 12b 12b 12b 12b 12b 12b 12b 12b 12	0C75 Mullard 25p 25 + 21p 100 + 17p	25 + 23p 100 + 20p 500 + 17p 1000 + 15p
AF115 25p BFX13 25p TIP33A AF116 25p BFX29 25p £1.00 AF117 25p BFX30 25p TIP34A	3 AMP PLASTIC WIRE ENDED RECTIFIERS 20+ 27p; 100+ 25p; 500+ 23p; 1000+ 21p. Any one type.	500 + 15p 1000 + 13p	0C72 Mullard 25p 25 + 20p
AF118 62p BFX37 32p £2:00 AF124 25p BFX84 25p TIS43 40p AF125 20p BFX84 25p TIS60 22p	PL7002 100 20p 19p 18p 17p 15p 15p All voltages 5:1-100V. 40p each. PL7003 200 22p 20p 19n 18n 16n 25+37p; 100+35p. Any one type.	0C20 97p Muliard 100v	100 + 17p 500 + 15p 1000 + 13p
AF126 17p BFX86 25p T1861 25p AF127 17p BFX87 25p T1862 27p AF139 30p BFX88 20p ZTX107 15p	PL7004 400 255 235 215 205 185 PL7005 600 250 250 250 250 250 250 250 250 250 2	25 + 85p 100 + 80p 500 + 75p	OC83 25p
AF178 47p BFY18 30p ZTX108 15p AF179 65p BFY50 22p ZTX300 12p AF180 52p BFY51 20p ZTX301 15p	PL7007 1000 30p 28p 26p 24p 22p Type Volts rent 1-49 50+ 100+ 100+ 100+ 100+ 100+ 100+ 100+	0C44 Muliard 17p	25 + 20p 100 + 17p 500 + 15p
AF181 42p BFY52 22p ZTX302 20p AF186 40p BFY53 17p ZTX303 20p AF239 40p BFY90 65p ZTX304 25p ASY26 25p BSX20 17p ZTX500 20p	(SILICON) SIZE † X † 1118. SCS5D 400 3 amps £1:00 85p 75p SC40A 100 6 amps £1:00 85p 75p SC40A 100 6 amps £1:00 85p 75p SC40B 200 6 amps £1:00 85p 75p SC40B 200 6 amps £1:00 85p 75p	25 + 15p 100 + 13p 500 + 11p	1000 + 13p
ASY27 32p BSX21 20p ZTX501 25p ASY28 25p BSX76 15p ZTX502 25p	1002 100 2 amps 60p 55p 50p 45p SC40D 400 6 amps 21.25 21.10 21.00 2002 200 2 amps 70p 65p 60p 55p SC45A 100 10 amps 21.25 21.10 21.00 SC45A 100 10 amps 21.25 21.10 21.00 800 75p 800 55p 8045B 200 10 amps 21.25 21.10 21.00	1000 + 10p OCI39 Mullard 25p	0C84 25p 25 + 20p 100 + 17p
ASY67 47p BSY95A15p ZTX504 40p ASZ21 42p BY100 15p ZTX531 30p	1004 100 4 amps 70p 60p 55p 50p 8045D 400 10 amps ±1.50 ±1.35 ±1.20 2004 200 4 amps 75p 70p 65p 60p 8C50A 100 15 amps ±1.65 ±1.55 ±1.35 4004 400 4 amps 80p 75p 70p 65p 8C50B 200 15 amps ±1.75 ±1.60 ±1.45	25 + 20p 100 + 17p 500 + 15p	500 + 15p 1000 + 13p
BA115 7p BY126 15p BA164 10p 6Y127 20p DISCOUNTS BAX13 6p BY182 85p 10% 12+ BAX16 7p BYZ70 40p 15% 25+	5004 600 4 amps 90p 80p 75p 70p SC40E 500 6 amps £1:50 £1:25 £1:10 1006 100 6 amps 75p 70p SC40E 500 10 amps £1:75 £1:50 £1:25 £1:10	OC81 Mullard 25p	AF239 42p 25 + 35p
BAX13 6p BY182 85p 10% 12+ BAX16 7p BYZ10 40p 15% 25+ BAY31 7p BYZ11 35p 20% 100+ BAY38 17p BYZ12 30p Anyone type	2006 200 6 amps 80p 75p 70p 85p 805 805 15 8050 500 15 amps \$2.25 \$2.00 \$1.75 4006 400 6 amps \$1.25 \$1.10 \$1.00 90p DIAC ST2 20p 6006 600 6 amps \$1.25 \$1.10 \$1.00 90p Larger quantity prices Evin. 4	25 + 20p 100 + 17p 500 + 15p	100 + 30p 500 + 25p 1000 + 20p
	-CONDUCTOR DEPARTMENT NEW LIST - NEW PRICES DISCOU	NTS We r	espectfully ask
_	WARE RD., LONDON, W2. TEL: 01-723 1008/9 LIST 1500 + 15% 25 ERS 356 EDGWARE ROAD W2. SEND FOR YOUR FREE 20% 11 POSTAGE	ANY ONE custo  + TYPE a min 10 + order 17 order 18 order 18 order 19 order	nimum £1,00 per ;it helpstoplanahear
	POŚTAG	ETP. ALLUNDERS & SE	ives postage as well.

Fig. 4: Wiring and layout of the audio amplifier board,  $3\frac{1}{2} \times 2in$ .

up as in Fig. 2, keeping all leads in this section short and direct. The metal flanged plate is the positive or chassis return, a tag being put under one fixing nut. Solder a black flexible lead to R3.

Tr1 is suspended by its wire ends about level with the top of L3 while trimmer TC2 is mounted on the paxolin by its tags.

IF Board Wiring. Fig. 3 shows both top and bottom of this panel. Holes are drilled so that the pins and can fixings tags on the i.f.t.'s are placed as shown. A central hole is required under the i.f.t.'s for trimming.

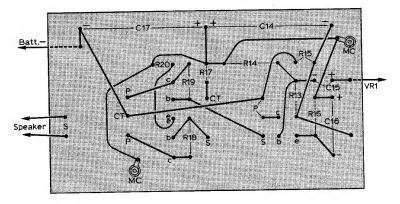
Two <sup>1</sup>2in. 6BA bolts secure the tags MC. Later, extra nuts are put on these bolts, which pass through the metal case, forming the positive or chassis

Guide lines are scribed on the paxolin and holes for resistors and other leads made with a 1/16in. or 5/64in. drill. The i.f.t.'s, resistors and capacitors are then inserted. Note the polarity of C8 and D1. The board is then turned over, and

wired as shown. Use insulated sleeving on all leads which cross. Joints are kept against the board so that they will not touch the metal of the case.

Before inserting the transistors, it is helpful to put pieces of 1mm coloured sleeving about 38in. to

T2 VR1 0 Batt. C14 C17



12in. long on the leads—blue for emitter, green for base, and orange for collector, leaving the shield wires bare. Then position the leads as shown in Fig. 3, solder them and snip off excess.

For easy identification of external connections,

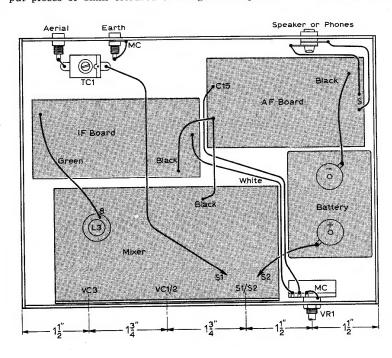
solder on a green lead for pin 2 of i.f.t.1, a black lead for the negative line, and a white lead from D1 positive.

Audio Amplifier Wiring. Fig. 4 shows both sides of the a.f. amplifier. Tags MC are tightly bolted as before, for chassis return points. All the capacitors are electrolytic, so have to be placed as shown.

T1 is the driver transformer but note that various makers place the tags in different positions. In Figs. 1 and 4, P-P are the primary, S-S the secondary, and CT the secondary centre tap. These connections must be correct for the actual transformer used.

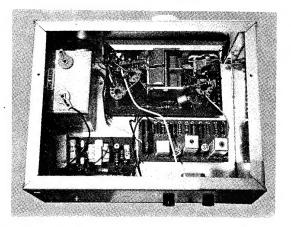
T2 is the output transformer, with primary connections P-P and primary centre tap CT. Two thin flexible leads are soldered to the secondary tags S which run to the speaker (or phones) jack.

Fig. 5: Disposition of the three boards inside the cabinet.



# \* components list

Resistors R1 10kΩ R11 4.7kΩ R16 11(0) 680Ω R9 2·7kΩ R7 68kΩ R12 1kΩ R17 8212\* R13 82kΩ 1kΩ R8 8-2kΩ R18 10kΩ\* 10kΩ\* R4 470Ω R9 680Ω R14 22kΩ R19 Ř5 39Ω R10 22kΩ R15 470Ω R20 4:7Ω All #W 10% except those marked \* 5% VR1 5kΩ potentiometer, log. Capacitors C1 22pF SM C7 330pF SM C13 0-01µF tub C2 470pF SM C8 10µF 6V C14 100µF 12V C3 0-01µF tub C9 0-04µF tub. C15 6µF 6V C13 0-01 nF tub. 0.01µF tub C10 0.5µF tub. C16 60uF 6V C5 0.01µF tub C11 0.04µF tub. C17 C6 22pF SM C12 0.04µF tub. 100uF 12V VC1/2 365 + 365pF 2-gang Type OO slow motion (Jackson) VC3 50pF miniature airspaced variable (Jackson) TC1 60pF trimmer TC2 60pF trimmer Inductors: L1 Range 5 'Blue' Denco miniature L2 Range 4 'Blue' L3 Range 4 'Red' type for transistors, IFT1 Type 18/465 IFT2 Type 18/465 Denco IFT3 Type 14/465 Semiconductors: Tr1 OC170 Tr2 AF117 Tr3 AF117 D1 OA81 Tr4 OC81D Tr5 OC81 Tr6 OC81 Chassis: Universal Chassis side 5 🗴 2in. Universal Chassis members:-2 off 8 × 4in. CU56A 2 off 6 × 4In. CU41B 2 off 8 × 6in. plates CU178 All from Home Radio Miscellaneous: T1, OC81D/2  $\times$  OC81 driver transformer T2, 2  $\times$  OC81/3 $\Omega$  speaker transformer S1/2, 2 pole, 3-way, wafer switch. Paxolin or 'eyelet' boards,  $5\times2\frac{1}{2}$ in,  $3\frac{1}{2}\times2$ in,  $3\frac{1}{2}\times1$ in. Knobs, 1-off type KN84F and 3-off type KN84D (Home Radio), Feet 4-off type Z146 (Home Radio).



Transistor leads can be identified as mentioned, or sleeving put on the wires to avoid possible short circuits.

Sclder on a black flexible lead, fitted with a battery negative clip. Leave a short wire projecting at the point IF negative, and connect an insulated wire from C15 positive.

# FINAL ASSEMBLY

If preferred, the mixer, i.f. and a.f. sections can be wired together and tested before fitting them in the case. The positive circuits must then be temporarily joined with wire.

The case front is a 4 x 8in. flanged member and three holes, for VR1, switch and VC3 are made  $1^{1}$ <sub>2</sub>in. from the bottom edge. The hole for VC1/2 is 2in. from this edge.

The fixing nuts of the switch and VC3 are then removed, so that the bushes can pass through the case front. If necessary, put a washer or two on each bush, then fit the mixer assembly to the case front, and lock it in position with the nuts of VC3 and the switch.

The bottom is then fixed. It was placed inside the flanges, but could be outside. Take a piece of paper of the same size, and mark the positions of L1, L2 and L3. Also the fixing bolts of i.f. and a.f. boards, and the central holes under the i.f.t.'s. Drill or punch these holdes, which are to allow alignment when all boards are in position.

The bottom is then bolted to the bottom flange of the front, the sides and back of the case being added later.

The i.f. and a.f. panels are fixed as in Fig. 5 by using extra nuts and lock nuts on the  $^{1}2$ in. bolts. There should be plenty of clearance, but if any joints are too near the metal, cut pieces of card to match the panels, and fit these under the i.f. and a.f. boards.

The green lead from i.f.t.1 is then cut and soldered to pin 8 of L3. VR1 is fitted to the front, and connected to positive line, D1 and C15, and other connections are made as shown.

When the case is to be completed, this is done by adding the two sides, then the back. All are bolted together with the flanges provided.

# IF ALIGNMENT

The i.f.t.'s listed are supplied pre-aligned. This means that signals should be obtained through the i.f. amplifier at good strength. If no signals are obtained, it is better to look for a wiring error or other fault, rather than move the cores of the i.f.t.'s.

When a signal is present, carefully adjust each core for best results. A screwdriver blade can easily break this type of core, and should not be used. If a proper tool is not available, it is recommended that one be obtained from the i.f.t. maker—the "Denco" TT5 tool is suitable.

Signals for i.f. alignment may be from a signal generator, input being to the base of Tr1, or from a transmission, correctly tuned in. If adjustment is by ear, use a signal which is quite weak even with VR1 near maximum. Alternatively, set a  $10k\Omega/V$  or other high-resistance meter at its 10V range, and clip it across VR1, negative to chassis. Alignment is then for maximum voltage reading. Only slight re-adjustment

# SERVISING

# AN INTRODUCTION TO FAULT-FINDING

PART 5 H. W. HELLYER

A KNOWLEDGE OF JUST HOW STEREO SIGNALS ARE GENERATED AT THE TRANSMITTER AND DECODED AT THE RECEIVER IS ESSENTIAL TO AN UNDERSTANDING OF THE STEPS INVOLVED IN THE ALIGNMENT OF STEREO RECEIVERS

HETHER or not you live in an area of the country served by a stereo transmitter, your interest in this series of articles must also encompass stereo broadcasting. Quite apart from the hideous interference that spoils a.m. radio on the "broadcast" bands, especially during peak hours, the restricted frequency response that results from a necessarily narrow bandwidth takes away any pretension to quality reception. With f.m. broadcasting, we can achieve a wider bandwidth and more nearly approach hi-fi—whatever that may mean.

# FM MULTIPLEX— THEORY AND PRACTICE

Nowadays, very few mono tape recorders and hardly any mono record players are to be found in the shops. Stereo is the order of the day, even in a cheap radiogram where the loudspeaker placement quite obliterates any good effect channel separation

15kHz 1 23kHz

pilot tone

Left channel Adder and To FM ın transmitter subtractor Right channe L+R+(L-R) double sideband+19kHz pilot tone Sub-channe sideband modulator 38kHz Frequency doubler 19kHz pilot tone 19kHz generator L+R

may have given! Stereo radio is the only way onwards.

Here is where that ugly word "compatibility" takes the stage. So that users of the existing mono system could continue to listen; so that stereo broadcasts could be "slotted in" to the current framework of programmes without robbing some poor listener of his entertainment, a system had to be devised whereby the stereo signals were combined in some way with the mono signals and some form of switched selector devised to differentiate between them

Fig. 1 shows how this is done in the system employed in this country. Here, we have the encoder at the transmitter which takes the stereo signal, applies it to an electronic adder and subtractor, and produces a left+right and a left-right signal. The difference signal, L-R is the vital stereo information. The L+R signal gives the mono listener all he

needs (with, it must be admitted, a slight degradation of the signal-to-noise ratio... but this is seldom mentioned).

The L-R signal modulates a sub carrier, the frequency of which is 38kHz. This is obtained by doubling the 19 kHz pilot tone output of a high stability oscillator. The trick here is to amplitude modulate the 38kHz sub-carrier at the transmitter. The modulator suppresses the subcarrier so all that comes out, for application to the adder, is the upper and lower sidebands of the signal.

Fig. 1. (top) illustrates the number of stages involved in the formation of a stereo signal at the transmitter while Fig. 2 (below) shows the frequency spectrum of a multiplex signal.

67kHz

Also applied to the adder is the untreated L+R signal, and a 19kHz pilot tone, which is to be used as a reference at the receiver. The signal spectrum at the transmitter, Fig. 2, helps to make this plain. This is the true multiplex signal. Suppression of the sub-carrier has given the two signals, L+R and L-R, 90% of the 100% modulation, the remaining ten per cent being taken up by the pilot tone.

It should be remembered that when L+R is maximum L-R is minimum and vice versa; in this way, the best use is made of the modulation "space". If the sub-carrier had been left in, the available "space" would have been reduced to 50%.

# COMPATABILITY

Fig. 2 shows that the ordinary mono information is still available up to quite a high frequency, around 15kHz. The useful sidebands of the 38kHz subcarrier extend to well over 50kHz. So a flat frequency response is needed throughout the propagation chain. One aspect not always considered is the need for good phase relationship. As the L+R and L-R signals depend on cancellation (in effect, the +L and -L components must cancel in the righthand channel and the R components in the lefthand channel) the phase relationships of the various signals must stay correct. One reason, you may argue, for a good aerial, and not just any old piece of wet string that happens to receive a signal. As we shall see, there are other reasons, too.

From the adder, this composite signal is fed to the transmitter and is used to frequency modulate it. It is this f.m. signal we receive, in the 88-97MHz section of the f.m. band.

Without going too deeply into this matter, we can state that there are two completely different ways of achieving this stereo signal and more than

two ways of decoding it. Referring again to Fig. 1, we note that an electronic "mixing" system has been used to encode the stereo signal. A "switching" system could have been used as easily. Similarly, the decoder, in the receiver, can be the matrix type or the simpler switching decoder which is more often used.

# SERVICING PROBLEMS

Tuners are undergoing some revolutionary changes just at present. Gordon King has said, and will be saying, a few things about varicaps, and tuning devices. From the servicing point of view, the

problems that arise are the same as in other tuners: component failures, displaced components, causing changes in inductance/capacity, where leads and supports are significant at the frequencies in use: plus the inevitable bogey of mistuning. Plus—it must be said—aerial mismatching, inadequate aerials and troubles with the feeder and its connections.

Through the if changed the main consideration

Through the i.f. channel, the main consideration is to achieve the necessary bandwidth and broad tuning is the order of the day. Attempts to "peak up" the i.f. channel of an f.m. receiver will inevitably result in an impaired response. IC's are being widely used in modern i.f. strips, giving very high gain. Tuning problems remain much the same, but are practically lessened by there being, in general, less tuned circuits to align. Ceramic filters, too, have made our life a little easier in the service department, refusing to go off tune, except when externally abused.

# THE DECODER

So let us get on to the decoder. This can either be the sum-and-difference type of Fig. 3a or the switching type of Fig. 3b. In the first, a mirror image of the encoder at the transmitter, three filters split the signal into its components. The first selects the 19kHz pilot tone, which is passed on to the frequency doubler stage. The second selects the mono L+R signal and presents it to the matrix. The L-R signal is selected by the third filter and passed to the balanced demodulator. To the same point the regenerated 38kHz sub-carrier is also fed. Addition and subtraction in the matrix now produces replicas of the original left and right signals.

The switched type of decoder is more common (less complicated!) and so will concern us more. Fig. 3b shows its block diagram. Multiplexed signals from

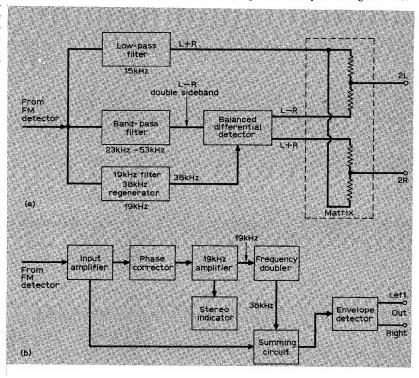


Fig. 3 (a) gives the requirements of a matrix decoder, the more common switching decoder being depicted in Fig. 3 (b).

the f.m. detector are fed to an input amplifier. From the output is selected the 19kHz pilot tone and the stereo information. No filters are necessary at the input to this amplifier, revealing at least one reason why the matrix decoder is out of favour in this country.

From the input amplifier, the pilot tone is phase corrected and amplified, then doubled in frequency to produce the 38kHz sub-carrier. An envelope detector is used, as with this method the left channel information is amplitude modulated on one side of the 38kHz carrier and the right channel information modulated on the other side. Alternate demodulation at a repetition rate of 38kHz gives a resultant audio output to each channel.

After demodulation, de-emphasis is applied. That bald statement is the reason why so many attempts at adding a decoder to a mono f.m. receiver to convert it to stereo failed dismally. Quite apart from the impaired S/N ratio—a factor which makes it vital to first obtain as good a mono signal as possible—the de-emphasis network which is fitted to counteract the high-frequency boost given to the signal at the transmitter has its effect as soon as we attempt to convert. De-emphasis is usually incorporated in the decoder design. The idea of this system is to obtain the best possible audio signal and also to attenuate residual sub-carrier components.

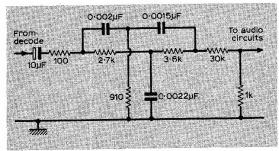


Fig. 4. Practical circuit of a 38kHz notch filter.

Despite this attenuation, some additional filtering may be needed if the stereo broadcast signals are to be applied to a tape recorder. Fig. 4 shows a notch filter suitable for removing all trace of the 38kHz signal from a tape recorder input. Many tape recorder manufacturers, particularly Continental ones, well used to f.m. reception, have already fitted filters, usually LC types in screened compartments near the input sockets. The circuit of Fig. 4 is easily constructed for experimental purposes by the person who adds his own decoder. It can give a fair measure of de-emphasis as well, and should be fitted between the decoder output and the amplifier input.

# DECODER CARE

Whenever any work is done at or around the decoder, connecting wires must be kept as short as possible. High frequency attenuation is easily caused by careless wiring, and instability by shoddy component placing. If a stereo socket is to be fitted, it should be mounted for efficiency rather than for convenience. There is always a temptation to cut corners—sometimes with costly results.

An example is the omission or removal of the

"cross-diodes" in a decoder circuit. Good stereo separation is still obtained; the signal is received clearly; there is no obvious interference. But more than once damage to output transistors has been caused, and it is possible that tweeters could be impaired in a hi-fi system, simply because the removal or failure of this circuit allows the 38kHz subcarrier to break through at full strength. We cannot hear it, but we shall certainly hear its effect!

One item noted on the block diagram of our decoder is the stereo beacon light. This is latched to the 19kHz amplifier and should light only when a pilot tone is present. Unfortunately, as any user of stereo receivers will know, it more often lights with every blip of noise.

# OTHER FACTORS

Other factors that are special to the f.m. receiver, and to stereo receivers particularly, are the automatic frequency control circuits, the muting circuits, and various devices to aid "capture." Some examples of these can be seen in Fig. 9 of Part 2 by Gordon King. (Page 136, PW, June 1971). See also the description of the a.f.c. circuits, on page 153 of the same issue.

Reference must also be made to my last part, where we began by treating the stereo receiver as we would a normal mono receiver, aligning for the correct response and checking the i.f. with a marker. The basic method is to connect the instruments as shown in my Fig. 6 of the last part, for normal f.m. alignment.

For more detailed work, a specially designed stereo generator is employed.

# TEST EQUIPMENT

An example of such an instrument is the Heathkit IG-37, costing £47·35 as a kit. This instrument provides a composite stereo signal plus a pilot tone and gives a phase test signal (i.e. channels added) for sub-carrier transformer adjustment. A variable r.f. oscillator signal is available, having an adjustable sweep width around a nominal 100MHz frequency. A mono f.m. signal can be obtained and this is modulated by any one of three modulation frequencies—giving a rapid "spot check facility."

In addition to this, four marker frequencies are supplied for r.f. alignment checks and two s.c.a. (subsidiary communications authorisation) frequencies for adjustment of filters. To the workshop that has to undertake regular checks of f.m. receivers, such an instrument can be invaluable.

# MULTIPLEX ALIGNMENT

With the proviso that the service manual for the receiver being aligned should be consulted initially, we now go on to detailed general steps of the alignment procedure of a typical stereo receiver, using the IG-37.

IF alignment: connect output of the generator, "RF Out," to the aerial terminals of the receiver. Connect the input of the last limiter circuit of the receiver to the vertical input of an oscilloscope, using a demodulator probe. Loosely couple the "IF Marker" of the generator to the first i.f. stage of the receiver.

This produces a 10.7MHz marker signal. Adjust the oscilloscope for a 60Hz line sweep. Use the "Mono/ FM" facility with audio modulation and tune the generator to approximately 100MHz. Then switch to "RF Sweep" input, with "IF marker." Produce a response curve (Fig. 5a) and adjust the successive i.f. stages for best response.

The next step entails comparison of curves and to do this we first need to view the generator output as applied to the oscilloscope and then beat in the signal from the receiver. To do this, first "kill" the tuner section, remove the vertical input of the oscilloscope from the receiver limiter circuit and connect it to the output of the detector stage.

Disconnect the i.f. marker. Connect the composite signal from the generator to the horizontal input of the oscilloscope, now set for external sweep, i.e. trigger the scope with the generator. With the tuner in action again, a curve as in Fig. 5b should appear. The detector coil can now be adjusted to approach this. Always refer to the maker's manual, if this is available.

Recheck the front end alignment using the r.f. marker signals at 90.95, 96.3, 101.65 and 107MHz adjusting the r.f. and oscillator coils for good response, and accurate tuning.

Stereo alignment: this needs a step-by-step operation, according to the type of circuit in use. In the IG-37 manual, these basic circuits are divided into four types (1) several 19kHz amplifier stages followed by a doubler; (2) a 19kHz oscillator circuit and doubler; (3) 19kHz amplifier, doubler and 38kHz oscillator, and (4) 19kHz amplifier and 38kHz oscillator. It is necessary to identify the circuit before proceeding, but some tests are common to all.

First set "Pilot Level" to minimum and function level fully clockwise, switch frequency to SCA and filter switch to 65 or 67kHz to suit the filter of the receiver, set function switch to "Audio" and "Mono/ FM," adjust the r.f. frequency adjustment to get a signal in a clear spot of the tuner band somewhere

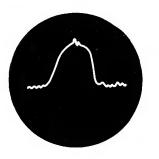
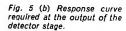


Fig. 5 (a) Response curve of i.f. channel of stereo receive with superimposed 10.7MHz marker signal.





near 100MHz. Make sure the phase control is central. If a separation control is provided on the generator. centralise this. Switch the a.f.c. off, and also, if available, the squelch circuit. Connect generator to aerial terminals. Switch on both generator and receiver and allow to warm up for 10 minutes.

If the set has an oscillator type of circuit, this should be "killed" for the SCA filter adjustment. If the filter is a complex or stagger-tuned one, the tuning will have to be done according to the maker's instructions. When feeding in the signal to the receiver keep the level such that the scope gives a clear trace-i.e., do not overdrive.

After this, proceed according to the type of circuit; the differences being the positioning of the oscilloscope, setting of the pilot level control and synchronisation of signals (19kHz and 38kHz) as

against peaking.

Taking the example of our Fig 3, i.e., 19kHz amplifier and doubler, we first peak both 19kHz and 38kHz circuits. The oscilloscope is connected to the output of the 38kHz doubler, pilot level turned to maximum (with regard to overloading), function switch set to Left channel, frequency switch to 1000Hz and the relative coils, including any stereo indicator coils, peaked for maximum indication.

Next, an operation common to all types, to adjust the phase of the reinserted carrier to the same phase as the sub-carrier. The oscilloscope is now connected to the left channel output, other settings are the same, and the phase is adjusted for maximum audio output. Normally, this operation requires the alteration of a 19kHz or 38kHz coil, and will be special to the particular circuit. Again, we need to refer to the maker's instructions, or trust our experience. No general guidance can be given without being misleading.

If a separation control is fitted, now is the time to adjust it. With the input to the left channel, we now read off the output from the right channel and adjust the control for minimum. (This operation can be done with the aid of the B.B.C's late night test

signals.)

Last step is adjustment of the 19kHz and 38kHz traps. Here, the pilot level control is turned to minimum, the function switch is again set to "Mono" and the generator frequency to the appropriate 19 to 38kHz. The oscilloscope reads off the appropriate channel output and adjustments are made for a minimum. Any automatic switching circuit that is fitted must here be locked according to the maker's instructions.

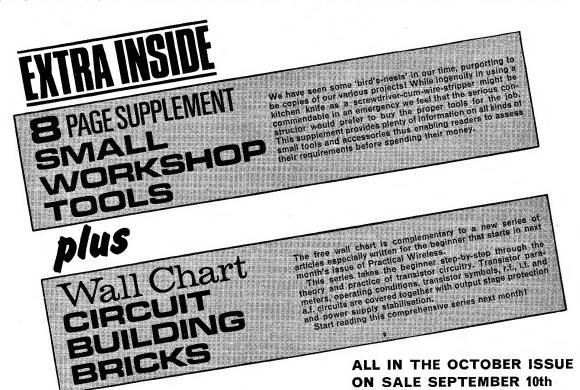
Now and again we come across circuitry that does not allow for easy adjustment. In these cases, always follow the maker's instructions explicitly. It is unfortunate that this has to be said so often. but circuits vary widely, as Gordon King has shown and which I hope my previous notes may have underlined. Something of the same conditions has to be met with audio equipment generally and with tape recorder circuits in particular, as we shall show in succeeding articles.

# END OF PART FIVE.

NEXT MONTH WE RETURN TO CO-AUTHOR GORDON KING WHO WILL DISCUSS THE DESIGN, CHARACTERISTICS AND PROBLEMS OF THE AUDIO SIDE OF EQUIPMENT BOTH VALVED AND SOLID-STATE

# \* IN NEXT MONTH'S

# WIRELESS



# **MODULAR AUDIO MIXING SYSTEM**

Consists of matching pre-amplifiers, tone control unit and line amplifier, etc., from which to build almost any audio signal mixing system from a simple three or four channel mono mixer for tape recording to a multi-input stereo signal mixer for discotheque or public address and music amplifiers.

The pre-amplifiers, seven in all, cater for:
Microphones with built-in transformers, high and low
impedance microphones, magnetic, ceramic and crystal
pick-up cartridges and guitar pre-amplifier (medium to
high Z).

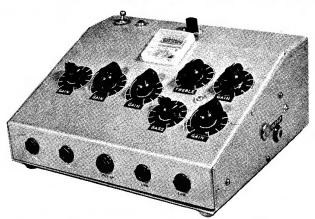
Each signal pre-amp has a built-in gain control and provides a 100mV output signal to match the bass and treble unit or the line amplifier unit.

The line amplifier provides up to 1 volt r.m.s. signal output suitable for driving power output modules direct, or any large external amplifier or a tape recorder.

large external amplifier or a tape recorder.

The system allows for the inclusion of line signal level inputs, from about 200mV to over 1 volt, which do not require pre-amplifiers and provision has been made for built-in VU meters and a suitable mains power supply. The system can be run from batteries for portable use.

The performance of the pre-amplifiers, tone control unit and line amplifier are to high fidelity standards. The various pre-amps, the tone control unit and line amplifier are simple to construct and can be panel mounted with one hole fixing. Full constructional details start in the October issue.



Plus many other constructional articles and all the regular features. Be certain not to miss the next issue, price 20p.

In the past, as far as the amateur or experimenter was concerned, there have been two main instruments for measuring frequency, the absorption wavemeter and the heterodyne frequency meter.

The absorption wavemeter has the advantage of being simple and responding to the fundamental frequency, but having a measurement accuracy limited to about 1%. It consists of an LC tuned circuit where the variable capacitor has a calibrated dial. It must be coupled to the circuit containing the frequency to be measured and when tuned to this frequency the absorption of power from the circuit may be noticed either by an effect produced in the circuit itself, or by an indicator attached to the wavemeter as shown in Fig. 1.

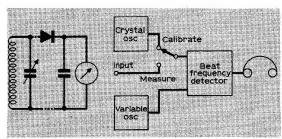


Fig. 1 (above left) Absorption wavemeter, Fig. 2 : (above right) Heterodyne frequency meter.

The heterodyne frequency meter is a much more sophisticated instrument which is capable of accurate frequency measurement. This is limited mainly by the accuracy of its internal standard oscillator and tuning dial. It consists of an accurate crystal oscillator, a stable variable frequency oscillator, with a finely divided dial and a beat frequency detector feeding headphones, as shown in Fig. 2. The variable oscillator is calibrated against the crystal oscillator by observing the zero-beat points where the fundamental or harmonics of the crystal and variable oscillators coincide. The input signal is measured by zero-beating the variable oscillator with it and then reading off and calculating the frequency from the crystal check points nearest to that of the input signal frequency.

Because the heterodyne frequency meter will also respond to harmonics of the input frequency, reasonable care has to be taken to identify and exclude these and other spurious responses.

# Frequency Measurement by Counting

An example of this method is in measuring the frequency of a heart beat. To do this, the doctor feels the patient's pulse and counts the number of pulses that occur in a fixed period of time.

The frequency of the heart beats can be stated in pulses per minute, no matter whether he counts for half a minute, one minute or ten minutes, simply by multiplying by a suitable factor.

Although electronic frequency counters have been in use professionally for many years, their construction has remained outside the range of home constructors, because of their complexity and high cost. However, the present availability of quite complex integrated circuits at extremely low prices has now made the construction of a frequency counter with digital read-out not only within the price range of

the home constructor but, in some aspects, its performance and price compares favourably with the cost of a heterodyne frequency meter available on the surplus market, such as the BC221.

# **Principle of Operation**

An electronic frequency counter consists of a switch, to define the counting period, in series with the input to an electronic counter and associated display, as shown in Fig. 3.

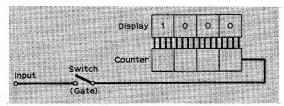


Fig. 3: Frequency counter and display.

Assume for a moment that the input signal is a square wave of frequency 1000 Hz. If we close the switch for exactly 1 second, then in that time 1000 cycles of the input signal will have passed through the switch to the counter. If the counter has 4 decades, comprising of thousands, hundreds, tens and units, then this will have completed a count of 1000 and as the counter is connected to a numeral display, the number displayed will be 1000.

It follows then that the frequency of any input signal will be indicated directly on the display, in cycles per second (Hz), providing that it is not greater than 9999Hz, which is of course the maximum reading for 4 digits.

# DIGITAL FRE

The digital frequency counter/ timer offers reliability, stability and read-out convenience, which can not be achieved using alternative techniques.

Wide use is made in this design of integrated circuits. Their use considerably simplifies the design, construction and adjustment. In its present form, this instrument is capable of operating in excess of 15MHz.

# JOHN THORNTON-LAWRENCE

Should we wish to make the same sort of measurement at 1MHz (1,000,000Hz) then we must arrange for the switch to be closed for exactly 1mS (1/1000 of a second). In that time 1000 cycles of the input signal will have passed through to the counter and the display will again show 1000, but this now represents 1000kHz, not Hz, (1000kHz=1MHz).

In a practical frequency counter, the switch is in the form of an electronic switch or "gate" and we call the time the switch is closed, the "gate time."

The accuracy of the frequency measurement made by counting depends mainly on the accuracy of the gate time. For instance when counting an input signal of lMHz for a gate time of lmS, an inaccuracy of  $1\mu S$  will cause an error of one count, which in the displayed answer represents lkHz. (There is also a counting error of one least significant digit, and this is discussed later.)

Below is a list of maximum frequency readings displayed by a 4 decade counter for different gate times.

Gate time	Maximum Reading
$100 \mu S$	99 · 99MHz
1mS	9.999MHz
10mS	999 · 9kHz
100mS	$99 \cdot 99 \text{kHz}$
1 Second	9.999kHz or 9999 Hz.

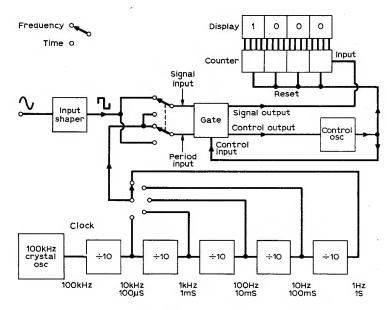
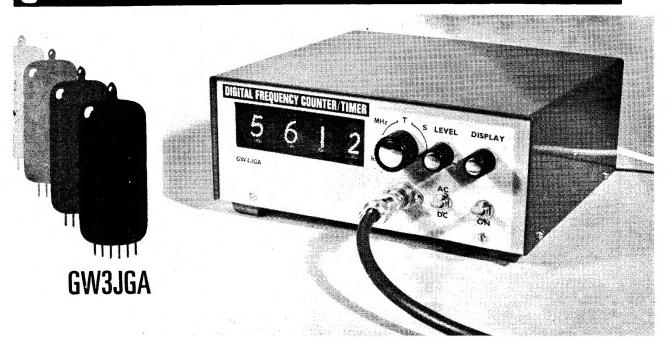


Fig. 4: Digital frequency counter timer.

To obtain the required order of accuracy it is necessary for the gate time to be derived from a very accurate and stable oscillator. A crystal oscillator operating at 100kHz and at normal room temperature 18°C-22°C, can be expected to have a frequency stability of about 1 part in 100,000, which is equivalent to 0.01% and is more than adequate for a 4 digit read out.

# QUENCY COUNTER/TIMER



The block diagram in Fig. 4 shows how various gate times may be generated and connected to the gate. In this arrangement the crystal oscillator operates at 100kHz and this frequency is divided or counted down in stages of ten giving the frequencies and gate times shown in the diagram. As this section determines the exact gate time we call it the "clock".

For example, with the clock switch in the 1 second position, the start of a 1 second square wave will open the gate (equivalent to closing the switch in Fig. 3) and the start of the next 1 second square wave will close the gate (equivalent to opening the switch), producing a 1 second gate time during which the input signal will pass through to the counter to be counted and displayed. It is assumed that prior to the gate opening, the counter is reading 0000; if this were not so then each successive reading would add to the previous one and defeat the object of the measurement.

# The Control Oscillator

To ensure that the counter is set to zero, a control oscillator is used. This oscillator has several functions; first it resets the counter to 0000, then instructs the gate to operate on the next clock signal and after the measurement has been made disables the gate for a few seconds, long enough to allow the displayed reading to be read before resetting to 0000 and repeating the operation. The sequence of operation is shown in Fig. 5 and will be discussed in more detail later.

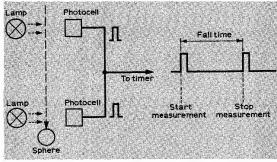
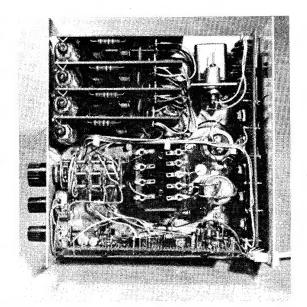


Fig. 5: Measurement of time between two events

# **Counter and Display**

The counter and digital display for a 4 digit readout consists of 4 identical counter and display units. Each digit has a decade counter which counts from 0 to 9, returning to 0 on the tenth count and providing a "carry" signal to the next higher decade. The counters count in binary form and the actual count figure is presented in binary coded decimal at 4 outputs, these having a weighting of 1, 2, 4 and 8 respectively.

The outputs are connected directly to a "decodedriver" which converts the binary coded decimal input to a decimal output. The 10 outputs are connected directly to a numerical indicator tube which displays the figure existing in the counter. A separate input connection to the counter enables it to be reset to zero at any time.



Photograph showing general layout of the digital frequency counter timer.

In the block diagram the input stage is shown connected between the input terminal and the gate. The purpose of this stage is firstly to shape the input signal to provide a square wave output suitable for driving the counter stages and secondly to provide a high input impedance.

# Time and Period Measurement

The basic component parts of a frequency counter can be rearranged very simply to provide the measurement of the period of one cycle of a low frequency signal or the time between two events. This is done by crossing over the clock and signal inputs by means of a change-over switch as shown in Fig. 4.

For example, suppose that we need to measure the frequency of a signal of 0.5Hz by the period method, (0.5Hz=2 seconds per cycle). With the change-over switch in the time position, the input signal now controls the gate and the counter counts the clock output.

If the clock output is 1kHz, that is 1 pulse per millisecond, and the period of one cycle of the input signal is exactly 2 seconds, then there will be a total count of 2000. We can convert this reading to frequency.

$$f = \frac{1}{t} = \frac{1}{2000mS} = 0.5000Hz$$

This means that we can now measure a low frequency with good accuracy.

The period measurement can also be used as an electronic stop watch for measuring the time between two events. For example, to determine the time taken for a sphere to fall a known distance, a set-up as shown in Fig. 5 may be used. Many other applications of this feature come to mind including race timing, camera shutter speed measurement, etc.

# \* components list

Resistors:	Semiconductors:
R1 $47$ kΩ R18 $150$ kΩ R2 $1$ MΩ R19 $1$ kΩ	Tri 2N3823 Texas Instruments
R3 2·2kΩ R20 150kΩ	Tr2 2N3702 Tr3 BSX20
R4 1kΩ R21 1kΩ	Tr4 BSX20 ,, ,,
R5 470Ω R22 3·3kΩ	Tr5 2N3704 ,
R6 680Ω R23 2.2kΩ	Tr6 2N3704 ,, ,,
R7 2-2kΩ R24 1kΩ	Tr7 2N3704 ,, ,,
R8 2·2kΩ R25 470kΩ R9 680Ω R26 82Ω 1W	Tr8 2N3704
R10 10Ω R27 82kΩ 1W	Tr10 2N3704 "
R11 220Ω R28 82kΩ 1W	Tr11 MJE521 Motorola
R12 1kΩ R29 82kΩ 1W	D1 1S44 Texas Instruments
R13 10kΩ R30 82kΩ 1W	D0 4644
R14 10kΩ R31 220kΩ R15 4·7kΩ VR1 50kΩ not carbon linear	D3 1S44
R15 4.7k $\Omega$ VR1 50k $\Omega$ pot carbon, linear R16 1k $\Omega$ VR2 50k $\Omega$ pot carbon, linear.	D4 1S44 , , ,
R17 1-5kΩ	D5 1S44 , , ,
all resistors 10% ‡ watt unless otherwise specified.	D6 VR525 (CV7172) A.E.I. (5-25V 1-5W 5% zener)
	D7 1N4007 Texas Instruments
Capacitors: C1 200pF S.M. 500V	D8 BS1100 p.i.v.1A
C2 0-47µF Polyester 400V	bridge rectifier   International Rectifier
C3 200pF S.M. 500V	D9 1S2051 Texas Instruments
C4 0-01μF Disc cer 15V	Indicator Tubes:
C5 50µF Elec 15V	V1 XN13 Hivac
C6 10µF 4	V2 XN13 , alternatively the XN3
C7 0-47µF Polyester 125V C8 470pF S.M. 500V	V3 XN13 , may be used
C9 60pF Concentric trimmer	V4 XN13
C10 50pF S.M. 500V	V5 Min. wire end neon. Radiospares
C11 1000pF Polyester 125V	V6 11 11 11 11 11 11 11 11 11 11 11 11 11
C12 $32\mu$ F Elec $450V(or16+16\mu$ Fin parallel)	
C13 100μF Elec 25V C14 2000μF μ, can μ, 1in, dia.	Transformer: T1 Midget mains trans. Radiospares.
C14 2000μF ,, can ,, 1in. dia. C15 0·01μF Disc cer 15V	T1 Midget mains trans. Radiospares. sec. 250V 25mA and 6-3V 1-2A
C16 0-01µF , , , , ,	
C17 0:01µF , , , ,	Switches:
C18 0.01µF :	S1 SPST miniature toggle S2 4 pole 11 way, built from:—
C19 0-01µF ,, ,, ,,	1 – miniature 'Maka-Switch' shafting assy.
Integrated Circuits:	4 – M-B wafers, 1 pole, 12 way
IC1 SN7473N Texas Instruments	2 – packets miniature spacers—Radiospares.
IC2 SN7400N	S3 DPDT Miniature Toggle
IC3 SN7490N , , , , , , , , , , , , , , , , , , ,	Crystal:
ICE ENTHON	X1 100kHz STC 4044 Type A
1C6 SN7490N , ,	Senator Crystals or Henry's Radio Ltd.
IC7 SN7490N ,, ,,	Metalwork:
IC8 SN74141N , , , ,	Complete metalwork (undrilled) in plain 16 s.w.g.
IC9 SN7490N ,, ,,	aluminium is available from H. L. Smith & Co.,
IC10 SN74141N ,, IC11 SN7490N	289 Edgware Rd., London W.2., at £1.40 (inc. p &p ).
IC12 SN74141N	Miscellaneous:
IC13 SN7490N , , ,	Veroboard 17-9 × 4-7in., 0-1 × 0-1in, matrix (part no.
IC14 SN74141N , , ,	VB126) cut to make 4 panels 2:325 × 4:1in., 1 panel
IC15 SN7490N ,, ,,	2.325  imes 5.5in., 1 panel $2.325  imes 7.1$ in. Crystal socket
I.C. Sockets:	10XJ, Brackets 12-LK2331 Lektrokit (Home Radio).
1.C. Sockers: 14 pin—dual in-line-11 off	Knobs with pointers. Filter, red 1½ × 4½in. Coax. socket, 3-3mm. insulated sockets, wire, nufs, screws,
16 pin—dual in-line-4 off	etc., 4 plastic stick-on feet, cable cleat, grommets,
Barnes or Texas Supply Co.	capacitor clips.

The instrument to be described operates according to the principles already outlined. Integrated circuits are used as far as is possible and these are of the Texas Transistor-Transistor Logic (TTL) 74 Series, which require a +5 volt supply and are capable of operating at frequencies in excess of 15MHz.

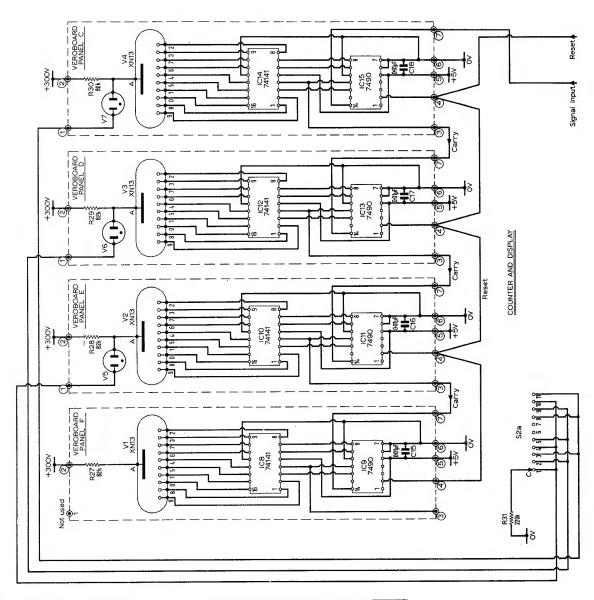
For this type of Logic the 'High' or logic '1' level is a voltage of greater than +2 volts and the low or logic '0' level is a voltage of less than 0.8 volts.

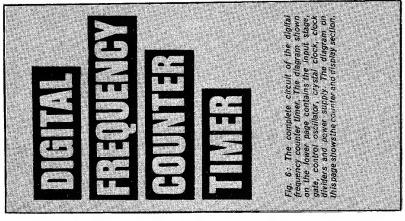
The numeral display employs Hivac neon numeral indicator tubes which operate from a 300 volt supply and give a clear bright display.

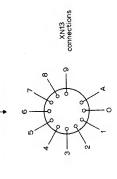
The clock crystal oscillator uses a 100kHz quartz crystal and this frequency and two of the divided frequencies are brought out to sockets for calibrating receivers, etc.

The complete circuit is shown in Fig. 6.

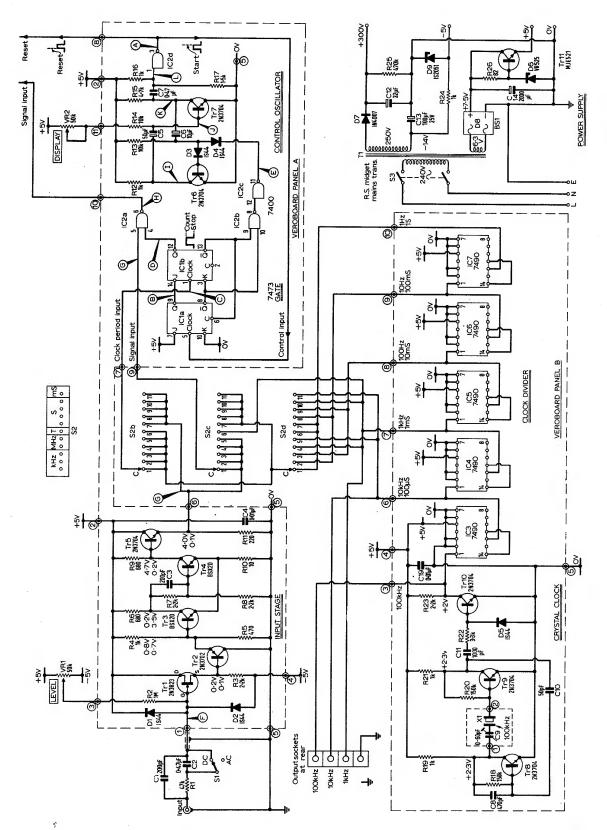
The input stage consists of an f.e.t. source follower directly coupled through a transistor emitter fol-

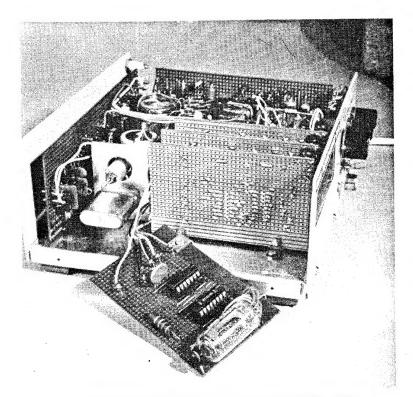




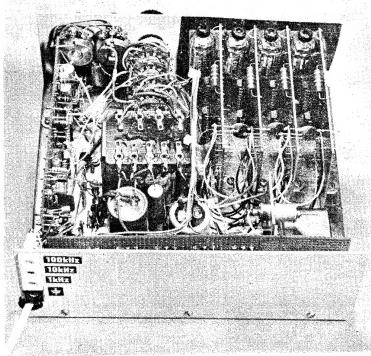


Direction of viewing





Counter and display panel removed, showing the integrated circuits and numeral indicator tube. The position of the oscillator crystal and concentric trimmer can be clearly seen at the rear of the counter and display panels.



Rear view of the instrument. The crystal oscillator and divider outputs can be seen on the rear metal panel.

lower to a Schmitt trigger circuit, which drives the TTL circuits through a further emitter follower.

Working backwards, the output from Tr5 must be sufficient to drive the TTL at frequencies up to about 20 MHz. This means that the output voltage must swing from greater than +2 volts to less than +0.8 volts at this frequency.

An input at the base of Tr3 of greater than +0.8 volts causes Tr3 to conduct, Tr4 to cut off and the output from Tr5 emitter to be +4.0 volts, representing a logic "1". An input to the base of Tr3 of less than 0.7 volts causes Tr3 to cut off, Tr4 to conduct and the output from Tr5 emitter to be +0.1 volts, representing a logic '0'.

For small input signals Tr1 and Tr2 have a gain of less than unity and an input of about 100 mV r.m.s. is sufficient to drive the Schmitt trigger circuit. To ensure that the output voltage at Tr2 emitter is at the correct level for switching Tr3, the gate of Tr1 is connected through R2 to a variable voltage provided by the 'LEVEL' control VR1. The LEVEL control is set at the position which gives reliable operation on small input signals. It will also allow the point of operation to be set to a particular level on a larger input signal.

For an input signal of a few volts peak-to-peak, Tr1 and Tr2 form an input clipping circuit having a high input impedance. At this signal level, Tr1 is cut off during the negative half cycles and Tr2 is cut off during the positive half cycles. This results in a square wave signal at the base of Tr3 having an amplitude which is more than sufficient to drive the Schmitt trigger reliably.

# Input Protection

The input to Tr1 gate is protected against excessive voltage by the clamping diodes D1 and D2. These diodes conduct at  $+5\cdot6$  volts and  $-5\cdot6$  volts respectively. R1 is included in the input circuit to limit the current that might be passed through the diodes, and C1 bypasses R1 for the higher frequencies. C2 is a direct voltage blocking capacitor which is in circuit in the 'AC' position of S1 and blocks off any direct voltage which may be present on the input signal.

In the 'DC' position of S1, C2 is shorted out and a direct input is available for very low frequency signals and for use in the timer or period mode of operation.

# Clock Crystal Oscillator and Divider

The clock consists of a crystal oscillator running at  $100 \mathrm{kHz}$  followed by five  $\div 10$  stages giving output frequencies and equivalent gate times as shown below:—

Frequency	Gate time
10kHz	$100 \mu S$
1kHz	1mS
100Hz	$10 \mathrm{mS}$
10Hz	100mS
1Hz	1 Second

# **Crystal Oscillator**

The crystal oscillator contains a 100kHz quartz crystal connected in the feedback loop of a two-stage amplifier formed by Tr8 and Tr9. Simple biassing is provided by R16 and R18. C8 is connected between the base and collector of Tr8 to provide the correct gain and frequency response for reliable oscillation of the crystal.

(N.B. A surplus 100kHz crystal was tried in the prototype and due to its lower activity it required C8 to be reduced to 200pF.)

The crystal operates in series resonance and C9 enables a fine adjustment of frequency to be made. The output from the oscillator is coupled by C11 and R22 to the clipping stage Tr 10. The square wave output at Tr10 collector has a suitable voltage swing for driving the first decade divider IC3.

Each decade divider consists of an integrated circuit decade counter connected to give a division of ten and a square wave output of unity mark space ratio. The outputs are fed to S2d and to the output sockets as shown in the circuit diagram.

# Control Oscillator and Gate

The gate actually carrying the signal through to the counter is IC2a. This gate is controlled by the Q output of the JK flip-flop IC1b which in turn is controlled by the JK flip-flop IC1a and the control oscillator.

The simplified sequence of events is shown in Fig. 7 and is as follows:—

The leading edge of a positive pulse from IC2d resets the coun-

ter to zero and the negative going edge sets ICla with Q high and  $\bar{q}$  low. This causes the output of IC2b to be high and the output of IC2c to be low, causing the clamping diode D4 to conduct. The control oscillator multivibrator is rendered inoperative by the conduction of D4.

The output condition of ICla presets IClb so that at the next negative going edge of the clock period input waveform Q of IClb will go high. This is the commencement of the gate period.

-continued on page 430

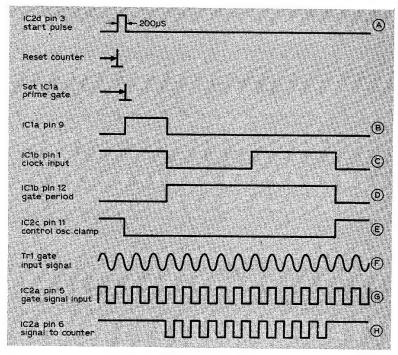


Fig. 7: Frequency counter gale waveforms

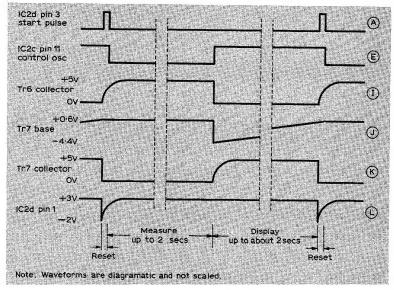


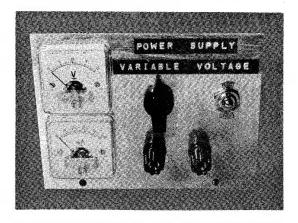
Fig. 8: Frequency counter control oscillator waveforms

# STABILISED POWER SUPPLY BILENGLISH

very popular piece of equipment in any experimenters workshop is a variable low voltage d.c. supply. Most transistorised equipment operates from supplies of from 3-30V and the unit described here was designed to give this output with a maximum current rating of 500mA which should satisfy the needs of most experimenters. An added advantage is that the unit is cheap to construct and rugged in use. Virtually any PNP transistors capable of withstanding 30V can be used in the unit and surplus transistors were found to give very satisfactory results. Tr4, of course, must be a power transistor capable of dissipating up to 10W, but no difficulty should be experienced if it is mounted properly on the aluminium chassis.

The theoretical circuit diagram is shown in Fig. 1. Full wave rectification is achieved by the diodes D1-D4 in the form of a silicon bridge rectifier type BY164 giving a d.c. voltage on C1 of about 22V. Two zener diodes in series provide the reference voltage on the base of Tr1. The voltage across these diodes remains constant at 20V over a wide range of current through the diodes. Consequently Tr1 will act as a constant current transistor with the potential developed across VR1 remaining fixed. Voltage control is achieved by varying the current through Tr4 connected in series with the load, this is done by VR1. Tr2 and Tr3 connected in the super-alpha configuration offer a high input impedance and prevent loading of VR1.

The prototype was constructed on a small tag board with flying leads to the transformer, power transistor, and output circuit. Details of the mounting and general layout are given in Fig. 2.



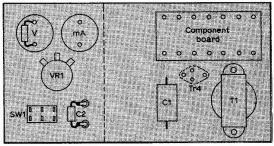
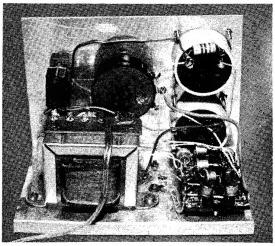


Fig. 2: A suggested component layout.



A rear view of the prototype. 0-500mA mΑ R1≷ 220≥ Tr3 AC128 AC128 кз≷ 2500µF C2 100µF ٧ 10-20V Fig. 1: ≷κ<u>.</u> 15k R2 The circuit of VR1 50k the power supply.

# THE BROADCAST BANDS Malcolm Connah

Frequencies in kHz Times in GMT

# MONTHLY **NEWS FOR** DX LISTENERS

THE recent bad weather has probably been the cause of the large number of reports received this month. The first report this times comes from Martin Ward in Portchester. Martin's equipment consists of a Mullard 5-valve domestic receiver and a 75-foot end-fed antenna, his log included: -

7285 R. Warsaw, Poland with music at 1200.

9620 R. Yugoslavia in English at 1530.

9750 R. Pakistan, Karachi in English at 1945.

11920 R. Kuwait with music at 2000.

15105 WIBS, Grenada with music at 2000.

15250 R.S.A., South Africa in English at 1800.

17845 R. WNYW, U.S.A. at 1700.

The next report is from Alan G. Crookes of Sheffield who has a Veritone CR-150 and a 200-foot long-wire at 20 feet, these enabled him to hear: -

6115 R. Berlin Int. in English at 2035.

6135 R. Free Europe in English at 2315.

9600 R. Baghdad, Iraq in English at 2030.

11860 R. Vilnius, Lithuania at 2335.

15160 R. Ankara, Turkey noted in English at 2205.

15345 R. Kuwait in English at 1800.

21520 SBC, Switzerland from Berne at 1530.

21535 R.S.A., South Africa at 1600.

Stephen John Mathews of Hull has used a Bush 4-valve domestic receiver and a 90-foot wire in his loft. This simple equipment brought in many interesting stations including:-

11672 Radio Pakistan in English at 0030.

11730 Radio Kiev in English at 0030.

11790 Radio Australia in English at 1710.

11760 R. Havana, Cuba in English at 0355.

11795 R. Nacional de Rio de Janeiro in Portuguese at

11805 R. Globo, Brazil in Portuguese at 0145.

11825 R. Jornal de Comercio, Brazil at 0040.

11914 Radio Nacional Lima, Peru at 2150.

11935 R. Clube Paranaense, Brazil at 2025.

15265 R. Afghanistan in English at 1810.

15290 Damascus Radio, Syria in English at 2040.

17765 VOA Tinang in English at 1700.

17855 NHK, Japan in English at 1020.

Hugh Cocks of Mayfield used his Unica UNR-30, 50-foot long-wire and outdoor TV aerial to compile an extremely long log which included:-

6135 HCJB, Ecuador at 0845.

6540 Pyongyang, N. Korea at 1900.

7235 R. Australia, signing off at 1730.

7270 R.S.A., South Africa at 2000.

9525 R.S.A., South Africa at 2100.

9550 R. Australia at 1700.

9580 BBC, Ascension Island at 0715.

9745 HCJB, Ecuador at 0930.

9912 All India Radio at 2000.

10040 Hanoi, N. Vietnam at 2000.

11675 Radio Pakistan at 2000.

11775 TWR, Bonaire, signing on at 0730.

11795 WINB, U.S.A. at 2100.

11835 Algeria in French at 1200.

11860 BBC, Ascension Island at 0700.

11910 ETLF, Ethiopia with news at 1930.

11955 BBC, Malaysia at 1645.

11960 VOA Thessaloniki, Greece at 1630.

15018 Hanoi, N. Vietnam at 2000.

15110 WIBS, Grenada at 2000.

15125 R. Australia at 0700.

15155 ELWA, Liberia in Arabic at 1615.

15185 Lagos, Nigeria at 0700.

15200 Lagos, Nigeria with news at 1530.

15295 TWR, Bonaire in English at 2145.

15325 Rwanda relay of Deutsche Welle at 0630.

Jeffrey Malina, London N.4, has a Skyrover Mk. II receiver and a 110-foot long-wire which enabled him to hear:

7245 BBC, East Med. relay at 2115.

9009 Israel, Jerusalem, English at 2115.

9480 Radio Kiev, Ukraine in English at 1930.

9545 BBC, Ascension Island at 2100.

9625 Israel, Jerusalem at 2045.

15250 R. Bucharest, Rumania at 1300.

17775 R. Afghanistan in English at 1815.

17875 VOA, Monrovia, Liberia in Russian at 2115.

21535 NHK, Japan in English at 0800.

21545 R. Accra, Ghana at 1445.

Colin Beesley of Bristol is another reporter with a Unica UNR-30 receiver which he uses in conjunction with a 75-foot aerial to hear stations which include: -

9725 R. Sweden in German at 1200.

9805 R. Cairo, Egypt with news at 2340.

11875 RAI, Italy in French at 1530.

11930 Radio Australia at 2230.

15160 R. Ankara, Turkey, news in English, 2300.

21545 R. Accra, Ghana, news in English at 1630.

25790 R.S.A., South Africa in English at 1625.

Christopher Gibbs in Camberley has a Trio 9R-59DS receiver, a 20-foot long-wire and a Joystick antenna which enabled him to hear:

5035 R. Clube de Cabinda, Angola with music at

6025 R. Portugal with DX programme at 2100.

6090 R. Luxembourg at 2109.

7210 R. Norway noted at 1805.

7220 R. Budapest, Hungary at 2132.

9525 R. Warsaw, Poland with news at 2030.

11755 R. Finland with news at 1815.

11940 R. Bucharest, Rumania in French at 1900.

Reports should arrive by the 15th of the month, and be addressed to the author at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.



# THE AMATEUR BANDS David Gibson, G3JDG

# Frequencies in kHz • Times in GMT

THE short skip conditions of the last month seemed to suggest that two metres should be showing some interesting results. However, logs from listeners did not bear this out. The skip was particularly noticeable on twenty where locals and near Europeans were coming in at colossal signal strength despite many of them using verticals, including some ground planes, which are often described in the antenna textbooks as just the thing for cutting down the local signals and picking up the DX

Conditions have been about average for the time of year with most of the DX packing into 14 and 21MHz. The lower frequency bands have not seemed to be too keen to part with any DX, at least, at this scribe's QTH. Ten metres has been up and down with the "down" being the more dominant of the two. Trouble is with ten, the moment you switch to another band the 28MHz sector can quite easily jump to life and permit the keen listener to log five continents in about the same number of minutes.

Topband has been lively as far as G stations are concerned with the usual crop of mobiles (including G3JDG) putting up quite a bit of r.f. Apart from the odd OK this band appears to have become a G-band unless someone has been hearing things I haven't?

John Moore (Leicester) says that his c.w. is quite good and claims that not only does he now find c.w. DXing as easy as phone, but he has the extra pleasure of knowing that fewer listeners are hearing his best c.w. DX stations. Just goes to show what can be done with a bit of persistence and determination. John bagged a couple of OK's on Topband and even managed to log s.s.b. squeaks from: MP4MBC, UF6FAX (c.w.) ZC4MU, ZS4RO, ZS6AYW, 5H3LV, 9J2RA, 9J2TF on ten metres.

"There's no 'H' in Witney," says **Stefan Kaye** (Witney with no 'H'). A peep at 80 metres with an AR88D and a 250ft. long wire at ground level revealed: CT1UEG, EA4ITU, EP2BQ, PY1AJ, VE1AGH, ZL4JF/A, ZL3LE, ZS1MH, all s.s.b.

T. Wright (High Wycombe) seems to bear out my remarks about 160 by saying that there have been "hoards of G stations about this month". Tim's log for 14MHz received on an R107 and 120ft. end fed includes: JA3EP, JX8IL and KP4GM plus a long list of EU stations. Looks like the short skip was quite prominent for best part of the month.

**D. Palmer** (Bishopbriggs) observes that I do not appear to receive many logs for forty metres. Receiver is a modified 19 set with a 7MHz dipole and a.t.u. which raised thirteen PY stations plus: CX1AA, CX1BBR, JW7UH, JW9QH, K1GZL, LU8AJG, VP2AA, ZB2A, 4Z4DX.

M. Bradford (Edmonton, N.9.) says that the l.f. bands have failed to produce anything interesting and that ten metres has been disappointing. This was not the case with 14MHz which obliged with s.s.b. from: CN8MC, CT2BB, EA8DJ, K4RON, KG4AL,

KG4EQ, LU4ECO, PY4AEB, VK2NN, VK5AZ, VP2BGL, W3UBM/MM, 4X4AE, 4Z4JW, 9K2YG. A listen on 21MHz produced: CE4ME, EA8GZ, JA4JBP, JA7EA, KZ5AA, LU2ECS, VS9MB, VS9MT, 9Q5OA.

Les took his homebrew v.h.f. Rx and it didn't work (hard luck Les), this is only part of the story. Three conspirators pooled their gear to form a combined assault on the amateur bands. The results on 20 were: FG7TD, HK1CDK, HK3AUE, HR1SO, HR5JDC, MP4BIN, T12AAC, UL7NW, VE1AVM, VE2DVV, VE3DLC, VK3BM, W5DRW, XE1WA, YV5BPG, and on 15 metres: CR6GA, W2AMM, W9IYY, YV1AVU, 4X4AE, 4Z4HF. Will the real D. Lawley, A. Wade and L. Allen please stand up? Incidentally, D. Lawley also sent in an individual log which included 27 VK stations received on his CR7OA and 64ft. end fed at about 18ft.

S. Lamprey (Cardiff) has a 9R59DE which has had a few unspecified mods done to it. The aerial is 30ft. of wire with a Joystick on the end. Signals received from: EP2SW, HV3SJ, JW5NM, JW7UH, JY1, M1ACH, MP4BJG, OD5FI, ST2SA, DJ3DH/TA, TU2AZ, 5Z4MI, 7Z2OM, 7Z3AB, 9K2AM, 9M2CP, 9Q5IA. No mode mentioned.

DX letter this month comes from **Tony Curtis** (Canberra, Australia). Tony is 17 and has been an s.w.l. for about four months. His s.s.b. log, using domestic receiver (Circa 1950), is fitted with a homebrew b.f.o. which has a bad attack of the drifts. However, the log for 20 metres reads: G6IA, G7LB, JA1EPJ, VK1EP, VK2ACD, VK3AJP, VK4AZ, VK5EB, VK6RU, VK8JS, VK9XI, countless W stations, ZL1ALW, ZL2ON, ZL3FM.

Happiness is JR-500S-shaped to **J. Iredale** (Llandudno). His pet is fed regularly with a dipole or 132ft. end fed (nothing like a varied diet) and in return for this loving care produced 20 metre evidence of: CT2AK, CT2BB, DU1DBT, JA3AAW, JA6AV, JA9YBA, JW5NM, K2LQQ/TF, KV4AM, KZ5JF, LU1FKF, LU8DFB, OX3BD, PY2CSV, PY3BXW, PY7YS, T12JO, T12WA, TG9LM, WA1NGK /TF, W6REH, ZB2A, ZL1PY, ZL2ABY, 4Z4HF.

A new CR7OA at the Coulsdon QTH of **J. Iggleden** is fed from a 215ft. network of wire (knit yourself an antenna kit?). Since installation it has produced 21MHz s.s.b. from: EA7DJ, JA8NJO, K5HYB, K6HTM, PY2ERS, TR2AA, W6ATW/M, W6TUQ, ZM6UP, 4X4AE.

Busy month for keen types. August 7-8, WAE c.w.; 8, Woburn mobile rally; 9, 2 metre s.s.b. contest; 15, 70MHz c.w. contest; 15, Derby mobile rally; 15, Torbay mobile rally; 22, Swindon mobile rally; 28-29, All Asia c.w. contest; 29, Stratford-upon-Avon mobile rally. September 4-5, v.h.f. n.f.d.; September 4-5, Region 1 IARU v.h.f./u.h.f. contest.

Logs, in alphabetical order please, to arrive by the 15th of the month to:

12, Cross Way, Harpenden, Herts.

# No. 83

# practically wireless HENRY commentary by

Odd Audio again–

THE more we get embroiled in that fringe area of wireless—what is now known as the hi-fi world—the less scrupulous salesmen seem to be. Trade Descriptions Act notwithstanding, some purveyors of audio equipment beg a great many questions when they prepare their sales brochures.

We are familiar with the 'what's a Watt?' argument, aware that output power can depend on all sorts of different measuring conditions, from the way you supply your amplifier voltage to the way you cock your head. By judicious juggling, one can make a humble ten-watter sound—on paper—like one of those monstrosities that pump P.A. around the local discotheque.

What we are not always ready, or able, to dispute, even if willing, is the more subtle business of distortion percentage, signal-to-noise ratio and overload. Even where these are stated, the trusting soul can be caught out.

Just lately, our illustrious contributor, Gordon J. King, has been insisting on applying more stringent tests. He would like makers to quote a true 'power bandwidth' and he requires more comprehensive distortion limits than a manufacturer is usually prepared to give. And what



Don't iump on me. Joe

Gordon wants, Gordon usually gets.

One of the things he wants is a statement of the output power over the rated frequency range of the equipment relative to the test frequency setting. The frequency limits where output falls by half then become the 'power bandwidth,' much more revealing than a single figure at a single frequency.

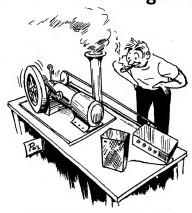
Moreover, he demands rms watts—none of your cheap and nasty Music Power for him. No use the amplifier designer arguing that real-life conditions do not encompass sinewave signals at full whack. Gordon and others argue that unless the rms test can be applied—continuous sinewave power—meaningful measurements cannot be taken and true comparisons can never be made.

Distortion is another matter for argument. A lovely low percentage at full power at 1kHz may make a Class B amplifier seem impressive. Check it again at 'normal' listening level, and see what you have got.

Then do a 'distortion bandwidth' test the way you did the power bandwidth test, i.e., over the specified frequency range. Oh boy! You should see the curves obtained from some quite well-regarded amplifiers when we make them suffer that indignity.

Some of the high-falutin' chatter that is going on about lownoise tape recording and the superiority of cassettes conveniently overlooks the fact that the distortion figure for a run-of-the-mill machine may be a couple of per cent—against the 0·1% we nowadays expect from a good amplifier and the 0·5% we usually get from a mid-price version. All right then—don't jump on me, Joe. I know that you cannot compute percentages like that.

But Henry would also like to see a regular set of standards applied. Even now, there is a committee advising another sub-committee, who are passing on their



Driving our amplifiers by steam

findings to the British Standards Institution, who may—eventually—stir their sluggish feet and complete BSS 3860; 1968 and 1568 and Part 1 1970.

Mind you, there is some very odd audio around. Take the Bose system. At 250-quid for a couple of speakers and an equaliser—you take it. Henry sticks his neck out in merely mentioning it for Dr. Aram Bose is currently sueing the American Consumer Association for more than they have got because they said his speakers needed tons of power to drive them and were not very effective when you did!

There is a review of the Bose 901 in the July 1971 issue of *Hi-Fi News*, where Ralph West keeps on about the 'smoothness' of the sound. He theorises about violins and Rolls-Royces being 'run-in' and the good sound of paper-cone speakers because the bugs are worn away.

Then the Editor chips in with a dry footnote to the effect that the original 4 inch drive units (Bose has nine of these per speaker) were taken from a Fisher FM table radio. You see? Odd audio or not, we come back to wireless in the end. Just think, if Benjamin Franklin hadn't earthed his key, we would have been driving our amplifiers by steam.

# **TAKE 2®**

# JULÍAN ANDERSON

A series of simple transistor projects, each using less than twenty components and costing less than one pound to build.

ELECTRONICS has brought about the introduction of excellent, inexpensive electronic flash guns. They are so cheap in fact that there is little demand for circuits to build one's own (though this is perfectly possible) but a number of ancillaries can be made for use with a flash gun.

One of the disadvantages with flash pictures is that the subjects often look harsh as narrow bands of dark shadow usually outline one edge of the subject. This, of course, is due to the flash point not being at exactly the point of the lens on the camera.

Bounce flash overcomes this but has the serious drawback that it is extremely difficult to calculate the exposure.

Another solution is to use two flash guns, one conventionally but a second one pointing from a very different angle whose job it is to soften the hard shadows and we can call this one a "slave flash". It would be possible to arrange some form of direct switching to trigger both from the shutter contacts but the one described here is completely unconnected electrically and is thus far more versatile.

The light from the master flash is arranged to trigger our little circuit, which in turn makes the contacts for the slave flash.

The reduction of components prices in recent years has made it possible for us to use some components that would not have been possible before in our series as the total cost would have exceeded the £1 limit. In this circuit we are using two such components: a light dependent resistor (LDR) and a thyristor or silicon controlled rectifier (SCR).

Light dependent resistors are exactly what their name implies. In bright light—such as produced by a flash gun—their resisitance is very low. The actual value varies considerably with the actual specimen but ranges between  $10\Omega$  and  $200\Omega$ . On the other hand in complete darkness a good sample will have a resistance of several megohms and even poorer ones at least  $10k\Omega$ .

This LDR is connected between the positive line and the base of Trl. In the ordinary way, even in complete darkness, this would be sufficient to bias the transistor into conduction, but VR1 is connected between the base and the emitter, forming a potential divider.

Being a silicon transistor, Tr1 will be off (i.e. no current will flow from emitter to collector) until the base is at about 0.6V above that of the emitter. It will thus be seen that, by correct setting of VR1, Tr1 can be arranged to be off in ambient light but as soon as extra light falls on the LDR, the potential at the base rises and the transistor is switched on.

The SCR in the collector circuit of Tr1 operates in much the same way as a relay. In normal conditions there is a very high resistance between anode

# No. 29 SLAVE FLASH TRIGGER

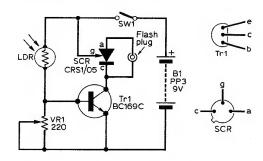


Fig. 1: The circuit of the slave flash trigger with the semiconductor connections shown on the right.

# \* components list

Tr1	BC169C	11pt
SCR	CRS1/05 Thyristor, 50V, 1A	25p*
LDR	Light dependant resistor	43p‡
VR1	220Ω linear potentiometer	12p†
SW1	On-off toggle switch	7 <del>½</del> p‡
		98½p

† Electrovalue Ltd.

\* G. W. Smith Ltd. or A. Marshall and Son.

‡ J. Bull (Electrical) Ltd.

Prices are those advertised in Practical Wireless July 1971 and may have changed. No allowance is made for minimum order costs or for postage and packing and these should be checked carefully before ordering.

and cathode but when triggered with a pulse at the gate, the resistance falls to practically nothing. In this circuit, for convenience, the gate is constantly at supply potential and it is arranged for the cathode to be made more negative, but it comes to the same thing as pulsing the gate.

If the electronic flash switch contacts are connected across the SCR, these will thus be shorted out when the light level on the LDR reaches a certain level. This operation takes place at electronic speeds and although there will be a delay between the main flash and the slave flash, it can be measured as a few millionths of a second.

Flash guns are usually fitted with 3mm coax plugs and the corresponding sockets can be obtained from most photographic shops, though the cost of this item takes us outside our £1 limit.

Note that the flash contacts *must* be connected the right way round, positive to anode, negative to cathode, and the potentials at the flash gun's plug should be checked.

As far as the camera aperture is concerned, this will depend on the film used and the direction in which the slave flash is pointed. Black and white film is not so critical as colour as far as exposure is concerned and no adjustment is necessary.

# 

ANY constructors will have in their spares box a selection of 1.4V filament battery valves taken from discarded "All Dry Portable" receivers. It occurred to the author that, when only intermittent operation over short periods is required, battery type valves of the B7G based miniature range form an excellent basis for the construction of simple test equipment. The one which forms the subject of this article was a wide range variable frequency audio oscillator, an extremely useful piece of equipment for testing audio amplifiers or the audio section of radio receivers. The majority of the old type battery portable receivers used a standard line up of four "DK" type valves, typically DK96, DF96, DAF96 and DL96. The requirements of the instrument described here are for two general purpose triode valves, and it was found that using the DF96 and DAF96 (or the older 1T4 and 1S5) connected as triodes, gave very satisfactory results.

The completed instrument will give a generous output (somewhere in the range of 2V peak) over a wide range of frequencies, from approximately 40Hz to 20kHz, the latter of course beyond the range of human hearing, but desirable when testing audio equipment. The coverage is given in three switched ranges (a fourth position of the range switch is the "Off" position), each of which slightly overlaps its predecessor. The waveform produced is good, and if an oscilloscope is available, the response of the amplifier under test can be accurately displayed and and tendency to distortion at certain frequencies becomes readily apparent.

# Circuit Description

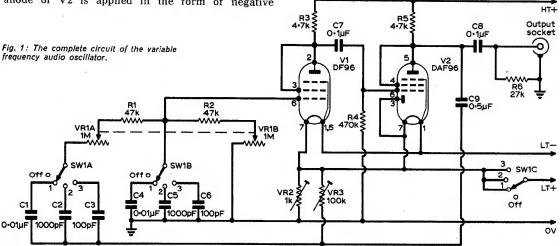
The circuit composes two triode connected valves, linked together by conventional RC coupling by the components R3, C7 and R4. The output from the anode of V2 is applied in the form of negative

feedback via the capacitor C9, and the potential divider network provided by VR2 and VR3 to the cathode (filament) of V1. Simultaneously, positive feedback is applied to the grid of V1 through the potential divider network comprising C1, C2 or C3 (as selected by the section "A" of the range switch) and VR1a, together with C4, C5 or C6 (selected by section "B" of the range switch) and VR1b.

Provided that the amount of positive feedback at the grid of V1 exceeds that of the negative feedback applied to its cathode, oscillation will take place and be maintained, the frequency depending on the setting of the range switch and the ganged potentiometer VR1. Both VR2 and VR3 are adjustable, and the circuit can be so "balanced" that steady a.f. oscillation is maintained throughout the range. The values specified for C1 to C6 are such that each "sweep" of VR1 provides the desired range of audio frequencies. It is important that VR1 is a dual ganged potentiometer of the linear type, otherwise a steady increase in frequency as the control is rotated clockwise will not be obtained. The third "pole" ("C") of the range switch is utilised to provide on/off switching, in the maximum anticlockwise position, the LT supply to the valve filaments is broken, and hence the instrument is "off", as no HT current can flow unless the valve filaments are heated by the passage of the LT supply through them.

The audio output of the intrument is picked off from the anode of V2 via C8, and taken to the coaxial output socket. In order to minimise the effect of connecting external apparatus of widely varying output impedance, a  $27k\Omega$  resistor is connected permanently across the output socket (R6).

-continued overleaf



# Testing the Instrument

Connect the output of the instrument to the input sockets of any available audio amplifier (valve or transistor) known to be in good working order, or to the pick-up sockets of a radio receiver. Set the fine frequency control to the mid-point of its range, and also set VR2 and VR3 to approximately midway setting. Switch on the amplifier (or radio receiver) and in the case of valve equipment, allow time to warm up. The volume control should be set approximately at the level required for normal radio (or record) reproduction. Now, having first made sure that the HT and LT batteries have been connected, turn the range switch to position 3, i.e., the middle range of frequencies. A note of approximately 400Hz should be heard in the loudspeaker. Adjust VR2 for maximum output, then reduce the setting of VR3 (by turning it anticlockwise) until oscillation ceases. Now re-advance VR3 until the oscillation just recommences. Next vary the setting of the fine frequency control VR1 and ensure that oscillation is maintained at all settings. VR3 should be set at the "minimum" consistent with maintained oscillation. Now set the switch to Range 1 (lowest), and finally to Range 3 (highest) frequency settings, and make sure that all positions of VR1 oscillation is maintained. Adjust VR3, and VR2 if necessary, to ensure that this is the case. The instrument is now completed and ready for use.

# TOPIC OF THE MONTH

"Electronic Vandals" continued from page 383.

why not join the society and take part in properly conducted explorations!

How does all this square up with the publication last month of the P.W. Treasure Tracer? In the first place let it be said that your editor is not only professionally involved in the making and using of electronic devices but is also an active archaeologist, a foot in both camps as it were. Basically it is felt that there is no reason why people should not enjoy innocent amusement with metal locators. It is generally thought that some site looters are unaware of the irreparable damage they can do so quickly, but a little thought will show that the robbing of archaeological sites is as indefensible as looting the showcases in a local museum. These sites are not only part of our heritage but also of future generations. So, don't get carried away with ideas of a vast fortune just waiting to be dug up-try the pools for that! And don't go tramping about on farmland without having first obtained permission from the owner.

W. N. STEVENS-Editor.

# **BACK NUMBERS**

We regret that the back numbers department of P.W. has now closed and consequently we are unable to supply these.

Requests for specific back issues can usually be included in our 'CQ' section; there is no charge for this but it is a service between readers and P.W. are unable to meet any of these requests.

## HARMONIC SIX—continued from page 398

of the cores should be needed.

Once the i.f. amplifier is aligned it should not be touched further as it is pointless to alter these cores again while aligning the mixer stage.

## MIXER ALIGNMENT

All the cores are adjusted so that about \$30\$ in. of brass rod projects. Screw TC2 about half down. Switch to the lower frequency range and tune in a signal with VC1/2 nearly fully open. Peak up the signal by rotating VC3. If VC3 is nearly fully open, screw down TC2 a little. With a signal peaked with VC3, tune towards the l.f. end of the band. When signals are tuned in, adjust the core of L2 for best volume. Continue until VC1/2 is nearly fully closed. If much adjustment of L2 core is needed, repeat from the h.f. end of the band. The purpose is to adjust TC2 so that signals peak with VC3 near the h.f. end of the band, with VC3 about half closed, then to set the core of L2 so that the minimum re-adjustment of VC3 is necessary, while tuning to the l.f. end of the band.

When this is done, switch to the h.f. range and repeat, adjusting the core of L1 near the l.f. end of this range.

If necessary, actual band coverage can be altered by moving the core of L3 near the l.f. end of the bands, and adjusting the trimmer TC2 near the h.f. ends of the bands. After any such change to coverage, adjust L1 and L2 to match, as mentioned.

# **AERIALS**

For a short rod or wire aerial, TC1 can be about half closed but for a longer aerial it should be unscrewed somewhat.

The aerial in use will influence the setting of VC3. Provided VC3 can be peaked for best results, at any frequency, and is not fully open or fully closed, there is no loss of efficiency due to wrong alignment. However, when alignment is correct, little or no adjustment of VC3 will be wanted while tuning, except perhaps to peak up very weak signals.

# NOTES

Resistors R4 and R5 are to prevent squegging and violent oscillation near the h.f. end of the oscillator range. With the coils and values shown, good results should be obtained without experiment.

In some cases a change in one or both values will be more suitable for the actual transistor fitted for Tr1. It may be found that R5 can be omitted, Tr1 emitter being wired directly to pin 5. With some transistors, R4 could also be omitted, or could be of lower value. R4 and R5 should be of the lowest values which do not result in numerous whistles and oscillation accompanying all signals tuned in near the higher frequency ends of the wavebands.

A 3-ohm speaker is suitable and should be in a cabinet, or fixed to a baffle board. A headset with headband is more convenient than a miniature or single earpiece for this type of reception. Medium or low impedance phones will generally be satisfactory.

The tuning dial may consist of a numbered dial, or two scales can be drawn on thin card, and marked with frequencies. The card can extend over the whole panel, and may be covered with thin transparent material.

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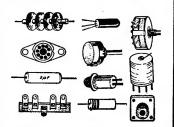
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# ELECTRONICS

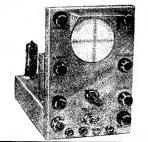
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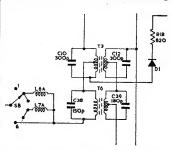
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# **EXPERIMENTERS CORNER**



# **NOVEL DOOR ALARM**

# **DAVID CROSS**

S the author's doorbell needed replacing, it was decided that an electronic replacement would be a good idea. It would be simple, but rather dull, to replace this by a simple audio oscillator feeding an amplifier and for this reason a two-tone version was designed, that is one that would provide two notes in rapid succession. Readily available components were used in the construction and on the prototype many of the components were found on discarded computer board panels.

The sound produced by this circuit has many uses other than for a door alarm and works equally well as a burglar alarm. This is especially useful as the sound produced can be made to sound almost exactly like that of a police siren—enough to worry any potential burglar!

When Tr1 is switched on, diode D1 conducts, this provides an additional conducting path for C2, thus shortening the time taken for the capacitor to discharge and trigger Tr2, which does exactly the same for C3 in the next half cycle. As it can be seen, if the resistance is high in VR2, VR3, the frequency is hardly altered, but as the resistance is decreased, C2, C3 alternately discharge through it, shortening their discharge times and therefore the frequency. The two pots used were actually salvaged from exequipment computer circuit boards, as were the transistors and some of the resistors.

The output from this multivibrator is picked up from the collector of Tr1 and is fed via C1 to the volume control of the amplifier.

The amplifier used by the author was taken from

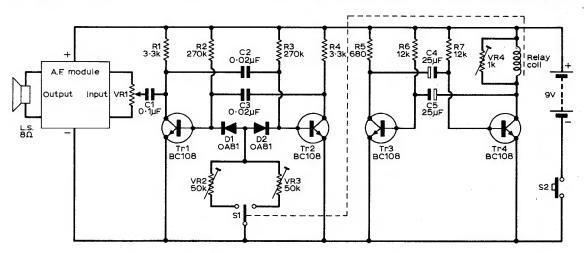


Fig. 1: The circuit of the two tone door alarm.

## THE CIRCUIT

The circuit comprises three main parts; two multivibrators and an audio amplifier. One of the multivibrators controls the rate of switching of the two notes, Tr3 and Tr4 provide this function and C4 and C5 control the rate of change. Tr4 has a reed switching coil in the collector circuit and a  $1 \mathrm{k} \Omega$  presentation of the two notes; it can be set for equal division or for any other spacing.

The second multivibrator is designed to operate at audio frequencies, the actual notes generated depend on the settings of VR2 and VR3.

an old battery tape recorder but any of the commercial amplifiers are suitable such as the Eagle EG 2004 250mW model. Many simple audio amplifier circuits have been published in Practical Wireless and these are also suitable.

Construction is a matter of personal choice but all the components can be mounted on plain Veroboard, the main wiring being provided using Cir-Kit strip. The existing bell wiring can be used and the bell push is represented by S2.

Once completed VR2 and VR3 can be adjusted for a pleasant couple of notes and VR4 adjusted for the balance between them.

# edium ave Column

HE new station at Kinshasa in the Congo has been widely heard in the U.K. with African style music and announcements in French. Listen before 0300hrs GMT while the channel is clear of QRM. On weekdays Deutschlandsender is off the air from 0130 to 0300hrs but on Sunday there is only a 15 minute break from 0245hrs. The north/south path is usually good during the summer. Gerry Wood from a QTH near Cape Town reports regular reception of the more powerful Europeans using a modified car radio.

Medium Wave stations in Newfoundland are often conspicuous in summer from 0200hrs GMT until sunrise. The strongest signal is from CBN 640kHz in St. John's which is the outlet of the Canadian Broadcasting Corporation. Satellites CBNA 610kHz in St. Anthony and CBNM 740kHz in Marystown usually carry the same programme which is announced as the CBC Radio Network. VOCM 590kHz in St. John's is the key station of a commercial network which is also represented by CKCM 620kHz in Grand Falls and CHCM 560kHz in Marystown. All three carry the same programme and have a common identification. CJOX 710kHz in Grand Bank and CJCN 680kHz Grand Falls relay the CJON Radio Service which has its chief station CJON 930kHz in St John's. Newfoundlanders are logged regularly in this country and are often noted

before sunrise as a cluster at the bottom of the band.

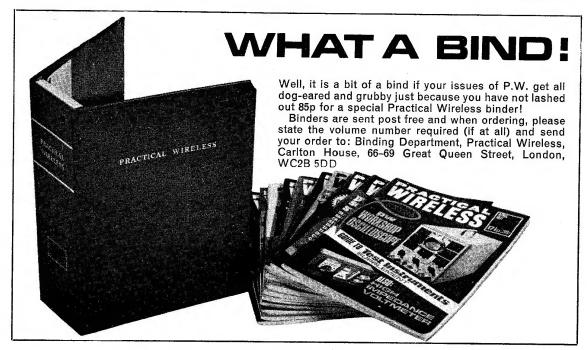
Broadcasting Stations of the World Part 2 is published at intervals of approximately 18 months by the Foreign Broadcasting Information Service of the United States Government. All known broadcasting stations in the long, medium and shortwave bands in the range 150kHz to 26MHz, in order of frequency, with some data on location, power and station identification, are listed except for MW stations in the U.S.A. The latest edition of the FBIS list, correct to January 1971, is now available. It can be obtained from the Superintendent of Documents, Government Printing Office, Washington DC20402, U.S.A., by quoting catalogue number PX EX 7.9.971 part 2, and sending a money order for \$2 in U.S. currency.

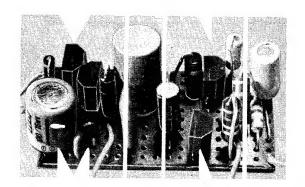
Leslie Smith of Witham, Essex, R. T. Johnston of Christchurch, Hants, and Frank Jeniker of Spur Tree Jamaica have written asking for information on MW loops. Anyone who would like to see an illustration of a loop should write to Radio Nederland and ask for the Frame Aerial Data Sheet. It can be obtained free of charge from Postbus 222, Hilversum, Holland.

Arthur Cushen of Invercargill, New Zealand, has recently verified his 2000th medium wave station. He sends the following details which will surprise many DXers. His QSLs are from stations in 132 different countries and 114 of them are from Europe and include every country. Arthur, who is blind, was awarded the MBE for his services to broadcasting. DXers, including the writer, who met him during his visit to this country in 1969 will be happy to congratulate him on his latest achievement. All reports and information concerning Medium Wave DXing may be sent to me at the following address:

Practical Wireless Editorial Department, IPC Magazines Ltd., Fleetway House, Farringdon Street, London, EC4A 4AD.

Charles Molloy





# amplifier

A LTHOUGH integrated circuits have largely taken over in the field of miniature electronics, it is still possible to construct a miniature audio amplifier which is economically competitive with an i.c. and which will give far more satisfaction to build.

The amplifier which is the subject of this article measures approximately  $2 \times 1 \times 1$  inches, and will deliver about 250 mW to a 250 m impedance loudspeaker. Its most obvious use is in the role of the audio stages of a small transistor radio, but it may

be used for many other purposes. It is possible to buy all the components (excluding the various miscellaneous items) for under £1.

# THE CIRCUIT

The circuit is given in Fig. 1. Four transistors are used, and these are all silicon types. Tr1 is an NPN, low noise type, which is used as the input stage. The input signal is coupled to the base of this transistor by C1, after having first passed the volume control. It is biased by the potential divider formed by R1 and R2, R3 being the load resistor.

C2 is a frequency correcting capacitor which is used to give negative feedback at high frequencies, where it has a lower reactance, and thus attenuating the higher frequencies.

Tr2 is a PNP type, and is used as the driver transistor. Biasing is performed by R4 and R5. As R5 is connected to the junction of the emitters of the two output transistors, as the voltage at this junction goes positive and negative with the input signal, the biasing of Tr2 will alter, and thus negative feedback is applied. R6 and R7 allow a small current to flow through the output transistors under quiescent conditions. This prevents crossover distortion which would otherwise occur as the input pulses change from one polarity to another. R6 and R7 also form the load for the driver stage.

# **OUTPUT STAGE OPERATION**

A complementary output pair are used, the BC159 being a PNP transistor, and the BC147 being the NPN type. These transistors are used as emitter

followers and so have a very low output impedance, and thus require no output transformer. The bases of these transistors are directly coupled to the driver transistor.

When a negative pulse is applied at the bases, the NPN transistor will shut off, and the PNP one will amplify, and its resistance will fall accordingly. This will cause the voltage at the junction of the two emitters to swing down to a very low level.

If a positive pulse is applied to the transistors, the PNP one will shut off, and the resistance of the

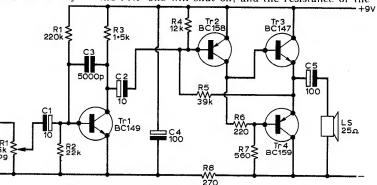


Fig. 1: The circuit of the Mini Amplifier. Note that Tr3 and Tr4 should preferably be matched transistors.

# \* components list

R1	220kΩ	R5	39kΩ
R2	22kΩ	R6	220Ω
R3	1·5kΩ	R7	560Ω
R4	12kΩ	R8	270Ω
All res	istors ±W, 10%	types.	
VR1	5kΩ log, pot	. with swit	ch
apacit			
	10µF 6V	C4	100µF 10V
C2	10//F 6V	C5	100µF 10V
C3	5000pF		
emico	nductors		
	BC149	Tr3	BC147
Tr2	BC158	Tr4	BC159
liscella	nanne		100
	pard, 2 x 1", 0 1	" matrix	
	Impedance lo		
	terv with batter		

-continued on page 428

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al	1000 pF	630	25p pack of 5		Working	Tolerance
Ĉ	1500 pF	630	<b>25p</b> pack of 5	Capacitance	Voltage	±20% Net
r	2200 pF	630	25p pack of 5	$0.01 \mu F$	100	27p pack of 5
$^{\mathrm{ls}}$	3300 pF	630	<b>25p</b> pack of 5	0.015 µF	100	27p pack of 5
Y	4700 pF	630	<b>25p</b> pack of 5	$0.022 \mu F$	100	27p pack of 5
Y	6800 pF	630	<b>25p</b> pack of 5	$0.033~\mu F$	100	<b>27p</b> pack of 5
	0.01 µF	630	<b>27p</b> pack of 5	$0.047 \mu F$	100	<b>27p</b> pack of 5
el	0·015 μF	630	29p pack of 5	0.068 µF	100	<b>29p</b> pack of 5
1.	0.022 µF	630	29p pack of 5	0·1 µF	100	<b>33p</b> pack of 5
	0.033 µF	630	33p pack of 5	0·15 μF	100	<b>38p</b> pack of 5
	0·047 μ.F 0·068 μ.F	$630 \\ 630$	40p pack of 5	0·22 μF 0·33 μF	100	<b>40p</b> pack of 5
	0.008 µF	63 <b>0</b>	<b>42p</b> pack of 5 <b>50p</b> pack of 5		100	44p pack of 5
	0·15 µF	630	<b>56p</b> pack of 5	$\begin{array}{ccc} 0.47 & \mu F \\ 0.68 & \mu F \end{array}$	$\frac{100}{100}$	52p pack of 5
	0.22 µF	630	67p pack of 5	1.0 µF	100	69p pack of 5 83p pack of 5
	470 pF	1000	<b>29p</b> pack of 5	1.5 µF	100	£1.23p pack of 5
	680  pF	1000	<b>29p</b> pack of 5	2.2 µF	100	31p pack of 5
	1000 pF	1000	<b>25p</b> pack of 5	$3.\overline{3}$ $\mu F$	100	41p pack of 5
	1500 pF	1000	<b>25p</b> pack of 5	4.7 µF	100	47p pack of 5
	2200 pF	1000	<b>27p</b> pack of 5	6.8 uF	100	<b>63p</b> pack of 5
				$0.022 \mu F$	250	27p pack of 5
	W	IMA TRO	PYFOI	$0.033~\mu F$	250	29p pack of 5
Э.	***	Working	Tolerance	0.047 μF	250	29p pack of 5
ņ	Capacitance	Voltage	$\pm 20\%$ Net	0.068 µF	250	33p pack of 5
d	47 pF	400	21p pack of 5	$\begin{array}{ccc} 0.1 & \mu F \\ 0.15 & \mu F \end{array}$	$\frac{250}{250}$	33p pack of 5
e	68 pF	400	21p pack of 5	0·15 μF 0·22 μF	250 250	42p pack of 5
0	100 pF	400	21p pack of 5	0·33 μF	250	<b>42p</b> pack of 5 <b>50p</b> pack of 5
	150 pF	400	21p pack of 5	0.47 µF	250	<b>56p</b> pack of 5
	$220~\mathrm{pF}$	400	21p pack of 5	0.68 µF	250	73p pack of 5
	330 pF	400	<b>21p</b> pack of 5	$1.0 \mu F$	250	88p pack of 5
	470 pF	400	<b>21p</b> pack of 5	$1.5 \mu F$	250	£1 25p pack of 5
	680 pF	400	<b>21p</b> pack of 5	$2 \cdot 2  \mu F$	250	<b>30p</b> pack of 5
	1000 pF	400	17p pack of 5	0.01 µF	400	27p pack of 5
	$^{1500}$ pF $^{2200}$ pF	400 400	17p pack of 5	0.015 uF	400	29p pack of 5
0	3300 pF	400 400	17p pack of 5	0·022 μF 0·033 μF	400	29p pack of 5
ė	4700 pF	400	19p pack of 5 19p pack of 5	0·035 μF 0·047 μF	400 400	31p pack of 5
n	6800 pF	400	19p pack of 5	0.068 µF	400	33p pack of 5 38p pack of 5
0	$0.01 \mu F$	400	19p pack of 5	0·1 µF	400	<b>40p</b> pack of 5
8	$0.015 \mu F$	400	21p pack of 5	0.15 µF	400	44p pack of 5
	0·022 μF	400	<b>23p</b> pack of 5	0·22 μF	400	54p pack of 5
	0.033 µF	400	<b>23p</b> pack of 5	$0.33 \mu F$	400	65p pack of 5
	0.047 µF	400	25p pack of 5	0·47 μF	400	81p pack of 5
_	0·068 μF 0·1 μF	400 400	27p pack of 5	0.68 µF	400	£1.13p pack of 5
3	0.15 µF	400	<b>33p</b> pack of 5 <b>44p</b> pack of 5	$\begin{array}{ccc} 1.0 & \mu F \\ 0.01 & \mu F \end{array}$	$\frac{400}{630}$	£1 36p pack of 5
	0.22 µF	400	<b>46p</b> pack of 5	0·015 μF	630	29p pack of 5 29p pack of 5
,	1500 pF	630	21p pack of 5	0.022 µF	630	33p pack of 5
1	2200 pF	630	21p pack of 5	$0.033~\mu F$	630	33p pack of 5
•	3300 pF	630	<b>21p</b> pack of 5	0.047 µF	630	38p pack of 5
•	4700 pF	630	<b>21p</b> pack of 5	$0.068~\mu F$	630	<b>44p</b> pack of 5
_	6800 pF	630	23p pack of 5	$0.1$ $\mu$ F	630	50p pack of 5
	0·01 μF 0·015 μF	630 630	25p pack of 5	0·15 μF 0·22 μF	630	56p pack of 5
	0.022 µF	630	<b>25p</b> pack of 5 <b>27p</b> pack of 5	0·22 μF 0·33 μF	630	71p pack of 5
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	1000 pF	1000	21p pack of 5	0.015 µF	1000	33p pack of 5
	1500 pF	1000	21p pack of 5	0.022 µF	1000	<b>36p</b> pack of 5
	2200 pF	1000	<b>23p</b> pack of 5	$0.033~\mu F$	1000	40p pack of 5
:	3300 pF	1000	<b>23p</b> pack of 5	0.047 µF	1000	44p pack of 5
L	4700 pF	1000	<b>25p</b> pack of 5	$0.068~\mu F$	1000	48p pack of 5
	6800 pF	1000	<b>27p</b> pack of 5	$0.1 \mu F$	1000	61p pack of 5
	0.01 µF	1000	27p pack of 5	0·15 μF	1000	83p pack of 5
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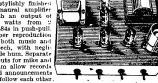
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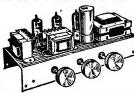
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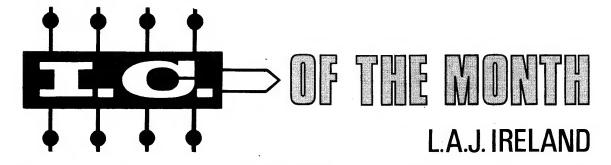
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Number 23

RCA CA3062 Photo-detector/amplifier

T the outset of the present series of articles it was pointed out that full constructional details could not be given in each article due both to the flood of new devices released on the market and the complexity of some of the i.c.'s reviewed. As a result some inexperienced constructors may have found themselves uncatered for as it would be too much to expect of them to draw up their own printed circuit and layout designs. This month, however, a very useful i.c. is reviewed and a layout pattern illustrated so that a relatively inexperienced constructor may approach it with confidence. Few external components are needed so that success is virtually guaranteed.

## Circuit

Photocell devices have always fascinated constructors as their uses are so varied and diverse from such things as counters and position sensors to the more familiar burglar alarm units. Considered in this month's article is the RCA type CA3062, described as a photo-detector and power amplifier and which is particularly suited for operations from either visible light sources or the newer infra-red Gallium Arsenide emitters.

Simple light dependent resistor control circuits suffer from two major drawbacks. First of all their response time is slow and secondly small variations in light intensity are not registered unless amplification follows. The photo-transistor is a much more

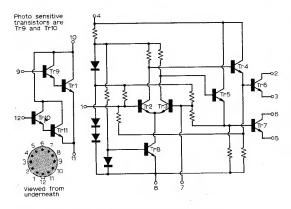


Fig. 1. Circuit of the photo-detector amplifier.

sensitive device especially when a pair of them are connected in a super-alpha configuration as in the present design. In fact the photo transistors themselves give a variation in current from a mere 4 microamps in the absence of light to 2 mA with a light intensity of 100 lumens per sq. ft. This current variation is usually used to develop a corresponding variable voltage across a 22kΩ load resistor which in turn controls the biasing on either Tr2 or Tr3.



This photograph of the CA3062 is 2½ times its actual size. The i.c. itself can be seen in the centre of the package which is a modified 12-lead TO-5 style.

In Fig. 1 a complete circuit diagram of the unit is shown and it can be seen that Tr2 and Tr3 form a differential pair with Tr8 acting as a constant current source. From this configuration it follows that phase inversion occurs between the collectors of Tr2 and Tr3 which in turn control the current in the output pair Tr6 and Tr7 via Tr4 and Tr5 which prevent loading of the differential pair. Some constructors may notice the great similarity in circuit design between this i.c. and an earlier one manufactured by the same company, type CA3020, and featured in the May 1968 issue of P.W. The main amplifier in both are almost identical except for the inclusion of a constant current sink Tr8 in the CA3062 replacing a simple resistor in the earlier i.c. to provide better stability and greater amplification. The same type of specially fabricated output transistors are used in each to enable them to carry the higher currents they draw as output transistors. In fact, up to 100 mA can be taken from the output stage and this is more than adequate to directly drive a relay or thyristor.

# Operation

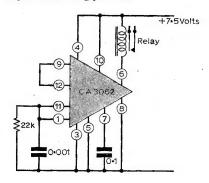
For proper on-off operation the voltage on either pin 1 or pin 7 (either may be connected to the photo transistors) should not exceed 3 volts. This can be achieved by limiting the current through the photo transistors or on the other hand by inserting a

clamping diode between whichever pin is used and earth.

It should also be noted that the device can also respond linearly to light variations by connecting a potential divider network between pin 6 and earth with the tap joined to pins 9 and 12 (two  $20M\Omega$ ). However, in this mode of operation the output current is limited considerably as greater heat dissipation takes place in the silicon chip.

# **Practical Circuit**

A suitable layout pattern for the photo-detector is shown in Fig. 2 and use could profitably be made of a TO-5 12 lead i.c. holder to prevent damaging the i.c. during the soldering process.



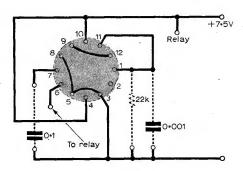


Fig. 2. The top circuit shows the two capacitors, one resistor and the relay needed to provide a practical circuit around the CA3062. The bottom diagram is a suggested layout for a circuit board using a TO5 12-lead i.c. holder.

In conclusion it may be pointed out that this device can provide light beam detection of up to 100 yards when used in conjunction with the new Helium-Neon lasers and which in the past year or so have appeared on the American market and no doubt will eventually filter over to this side of the Atlantic. The prohibitive prices of these devices severely limited their uses to industry and research centres but now one of these laser tubes sells for £20 and a suitable power supply to operate it can be built for under £5. No doubt this offers intriguing possibilities especially in the field of intrusion alarms and we can continue to expect in the future further developments along these lines.

# MINI-AMP-continued from page 422

NPN one will fall. This will cause the voltage at the junction of the two emitters to swing to a very high level.

In this way the emitter voltage will swing up and down with the input signal. This voltage swing is coupled to the speaker via C5.

Under quiescent conditions the voltage at the emitter junction should be approximately half the supply voltage. This voltage is set by the values of resistors R4 and R5.

If the amplifier is to work properly, the output transistors should be fairly evenly matched. These transistors are not generally available in matched pairs, but some semiconductor suppliers will match transistors, although there may be a small surcharge for this.

Current limiting resistors are often connected between the emitters of the output transistors and the output capacitor, to prevent thermal runaway. Silicon transistors, however, do not suffer from runaway to anything like the extent that germanium types more usually found in this type of circuit do, and these resistors were found to be completely unnecessary.

Although the amplifier will operate using speaker impedances outside the range  $15\Omega$  to  $30\Omega$ , this is not recommended as it will mean either a loss in available output power, or an increase in the level of distortion.

# CONSTRUCTION

The amplifier is built on a  $2 \times 1$  inch piece of 0.1 inch matrix Veroboard, with the copper strip running lengthwise. This has to be cut from a larger piece of Veroboard using a small hacksaw. A wiring diagram of the panel is given in Fig. 2.

Great care will have to be taken with the connections to the Veroboard, and this project should not be attempted by the complete beginner. A soldering iron with a very fine bit is required, and some sort of stand to hold the board steady will be found to be extremely helpful.

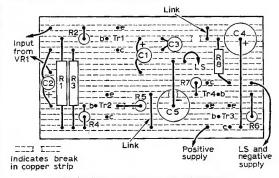


Fig. 2: The component layout on Veroboard.

All transistors are fitted with special leadouts which plug into the Veroboard. These leadouts are very short, and although silicon transistors are very hardy, these should be soldered quickly into place, and should be left until last.

The capacitors will have to be subminiature types, and like many of the components may have to be mounted vertically so as to fit into the available space.

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BP12174121	Monostable Multivibrators		770	67p
BP141 = 74141	Monostable Multivibrators		77p	67p
BP145==74145	BCD-to-Decimal Decoder/Drivers	1.80	£1.70	£-160
	8-bit Data Selectors (with Strobe)	1.40	£1.30	£1.20
	Dual 4-Line-to-Line Data Selectors/			
DY 100 - 14109	Multiplayere	1.40	£1-30	£1.20
BP101-7/101	Multiplexers	2.50	£3.25	£3.00
Davisse may b	a mixed to qualify for quantity price. Larger	anant	ities-	prices or
application /T	TL 74 series only).	quant		F 01

Data is available for the above series of I.C.'s in booklet form. PRICE 13p.

			IAPMII II P			
Type No.	Case	Leads	Description		Price	
				1-24	25 - 99	100
BP201C-SL201C	TO-5	8	G.P. Amp	63p		45p
BP701C-SL701C	TO-5	8	OP Amp	63p		45p
BP702C-SL702C	TO-5	8	OP Amp Direct O/P	63p		45p
BP70272702	D.I.L.	. 14	G.P. O.P. Amp (Wide Band)	53p		40p
BP709-72709	D.I.L.	14	High Gain OP Amp	53p		40p
BP709PvA709C	TO-5	8	High Gain OP Amp	53p		40p
BP74172741	D.I.L.	14	High Gain OP. Amp (Protecte	d)75p		50p
4A703C-4A703C	TP-5	6	F.RIF Amp	43p		27p
TAA263	TO-72		A.F. Amp	70p		55p
TAAA293	TO-74	10	G.P. Amp	90p	75p	70p

TTL INTEGRATED CIRCUITS Manufacturers' "Full outs' —out of spec, devices including functional units and part function but classed as out of spec, from the manufacturers' very rigid specifications. Ideal for learning about L.O.'s and experimental work.

See p

$\begin{array}{llllllllllllllllllllllllllllllllllll$							
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	PAK No.	PAK No.					
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	UIC00=12×7400N 50	p UIC42= 5×7450N	50p UIC80=				
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	UIC01=12×7401N 50	D UIC50=12×7450N	50p UIC82=	5×7482N 50p			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	UIC02=12×7402N 50	D UIC51=12×7451N	50p UIC83=	5×7483N 50p			
$\begin{array}{llllllllllllllllllllllllllllllllllll$	UIC03=12×7403N 50	UIC60=12×7460N	50p UIC86=	5×7486N 50p			
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	UIC04=12×7404N 50	UIC70= 8×7470N	50p UIC90=	5×7490N 50p			
$\begin{array}{llllllllllllllllllllllllllllllllllll$	UIC05=12×7495N 50	n UIC72= 8×7472N	50p UIC92=	5×7492N 50p			
$\begin{array}{llllllllllllllllllllllllllllllllllll$	UIC10=12×7410N 50	n UIC73= 8×7473N	50p UIC93=	5×7493N 50p			
UIC40=12×7440N 50p			50n UIC94==	5×7494N 50p			
UIC41= 5×7441AN 50p UIC76= 8×7476N 50p UIC96= 5×7496N 50p				5×7495N 50p			
			50n UIC96=	5×7496N 50p			
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$UIC935 = 12 \times \mu A 935$		50p	UIC9093 = $5 \times \mu A$ 9093	50р
$UIC936 = 12 \times \mu A 936$		50p	UIC9094= $5 \times \mu A$ 9094	50p
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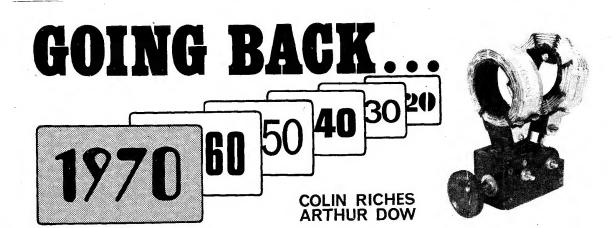
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POLLOWING our brief mention of the type LS3 valve in 'Going Back' for June we were pleased to hear from George Jessop, G6JP, who has been with one of the leading valve manufacturers of this country for many years. George points out that the LS family of valves was introduced to make available larger output powers than were currently available with the 'R' type valve which had first appeared around 1915 (see 'Going Back' July 1971).

LS' indicated a loudspeaker valve and applied to the earlier series LS1-6 and to several later types such as the LS7, 8, 8A and 9 although these had little in common with the earlier series, having oxide coated filaments and being specially manufactured for the Post Office.

The most famous valve of the early range was undoubtedly the LS5 which was for general purposes while the LS5A was meant for use as an output valve. The higher impedance LS5B completed the range and was intended for use in resistance-capacity coupled stages.

Both the Post Office and the BBC used these valves in their high quality audio amplifiers while, with the radio amateurs, the LS5B reigned supreme in frequency multiplier stages and the LS5 did duty in the power amplifier output stage. G6JP mentions that in 1936 his first 10-metre transmitter used these valves and that a very early 2-metre oscillator, illustrated in "Short Wave Communication" by Ladnor and Stoner (1932), used two 'de-based' LS5's. Since the circular base of the standard LS5 was of metal a considerable reduction in self-capacity was possible by removing the base thus enabling an output of some 5 watts to be obtained from this particular oscillator.

Readers may like to know that this famous range of valves has been graced with CV numbers:

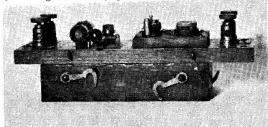
LS5	Metal 4-p	in IIX h	ase	VT24CV1636
200				VT25—CV1637
LS5	Standard	British	pase	
LS5A	,,	"	27	VT66—CV1650
LS5B			_	VT40—CV1647

All these valves had an anode dissipation of 10 watts but the later LS6A had this figure increased to 25 watts.

Our thanks to George Jessop for all this interesting information and we are sure that if he cared to put his reminiscences on paper they would prove very absorbing reading.

# Bintage Equipment

One of our correspondents has sent this photograph of an ancient piece of gear and asked us to assist in identifying it. Unfortunately we can't help so we are passing the buck to our older but nevertheless revered readers in the hope that someone can recognise the equipment.



Two of the terminals are marked 'aerial' and 'earth' and the other two 'cell'. A couple of coils, a buzzer and a knifeswitch complete the 'thing'. The switch looks rather like a Morse key and no doubt could be used as such with a bit of ingenuity.

Our correspondent did connect a 1.5 volt 'cell' as instructed and reports "it did no good to our TV set". We can imagine! We think it is a crude transmitter for test purposes which could be left running while a receiver was adjusted. Any other ideas?

# Devices

Assuming that one or two readers of 'Going Back' have been coerced into using solid state devices, they have probably endeavoured to follow the advice of their betters by using a heat-sink when soldering the leads of transistors and keeping the metal clips on certain FET's until they are safely tucked in. (The FET's, not the readers!)

However the early users of valves also suffered a barrage of injuctions on how to handle their new toys. The following directions, taken from the wireless press of 50 years ago, should raise a smile or two!

# BREVITIES FOR VALVE USERS

1. Remember that your receiving apparatus is not a crude collection of switches, but that the operation of every switch should receive due consideration

- 2. Turn on your valves slowly.
- Do not burn them too brightly.
- Signals are not necessarily stronger the brighter the valves are lighted.
- 5. Do not let your valves 'howl'. Your neighbour may report you and you risk having your licence cancelled.
- 6. Having roughly tuned in on your inductance and accomplished finer tuning with your variable condenser, bear in mind para. 5.
- 7. Remember there is a proper arrangement for the valves. Interchange them from left to right, having tuned in a station, such as FL, until you find the best position for each.

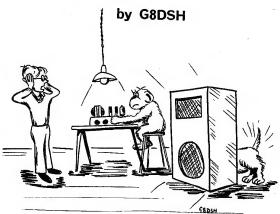
We are sure that reading these 'instructions' will bring back many memories to those who fiddled about in those pioneering days. At least one could see when a valve had failed which is more than one can do with today's 'devices'!



This year we are sponsoring another Project Autumn competition. The rules have been amended so that the PW "Designer's Trophy 1971" will be awarded to the author of the best constructional article published in PW issues dated July 71 to March 72 inc. This allows submitted articles to be published as soon as possible and for authors to be paid without delay.

SUBMIT YOUR ARTICLES NOW

# MAXWELL



"Your bass response is dreadful . . . whatever are you using for a Woofer?"



Items in this section are carried free of charge as a service to readers. We only ask that those making use of the service answer all correspondence resulting and reimburse postage and all reasonable expenses. We cannot guarantee inclusion and reguests for inclusion will not be acknowledged, Material for inclusion should be sent to Practical Wireless Editorial. Fleetway House, Farringdon Street, London, E.C.4.

INFORMATION WANTED
...circuit diagram for the BCC 715. All costs refunded.—Eskil Persson, Frotunagrand 1, 19400 Upplands Vasby Sweden.
...handbook for Admiralty B40 (Murphy) receiver.—M. Johnson, 5 Barons Court, Haverhill. Sufficial

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...hiotomation on the fitting of a magic eye tuning indicator or S-meter to a mains
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Street, Halfeld, Pretor his subject.—J. Towers, 107 Hatfield Galleries, 1068 Burnett
Street, Halfield, Pretor insured, South Africa.
...any mods. to the B40C receiver..—B. Eames, 38 Fairway Crescent, Allestree,
Derby, D53, 2PE.
...circuit or info. required for Tx/Rx type CR1—43044 (TBY8)—borrow or buy —D.
Byrne, Kylemore, Endsleigh, Douglas Road, Cork, Eire.
...titles of books on electronics for beginners. All correspondence to C. S. Powell,
30 St. Michaels Way, Brundall, Norwich, Norfolk, Nor8.82.
...Instructions or valve line-up for U.S. war surplus Valve Tester I-177.—Roger
Decamps and J. Dr. 274, 2800 Malines, Belgium.
Grantham, Lincs.
...address of place where I may get ferrite rod aerial for Bush mains radio AM82
or any gen on how to make same.—S. C. Moodley, 5 Dale Close, Addlestone, Surrey,
...information on adding a second manual to the Polyphonic Organ published by
P.W. some years ago.—R. Wendrick, 22 Longcliffe Gardens, Nanpanton, Loughonson's Radio, St. Martins Gate, Worcester.

porougn, Lincs. ... instruction manual or any gen on Cossor 1039M oscilloscope.—G. J. Banner, Johnson's Radio, St. Martins Gate, Worcester ... .. technical information required regarding ex-Army (Mazda) A720, CV1381 transmitting valve.—P. A.G. Price, 58/57 The Mall, Meerut, Cantonment, U.P., India. ... any information at all on the surplus W/S 31 especially the valve line-typ, supply connections and handbook.—J. H. McGee, 6 Ripon Road, Newton Hall, Durham, Co. Durham.

Durham. ....avy info. on converting the 19 set Rx/Tx.—T. Izard, 41 Clumber Drive, Radcliffeon-Trent, Notts. on-Trent, Notts.

on-Trent, Notts.

a fault-finding chart No. 2 which was issued with the May 1968 issue of P.W.—
Norman Quinn, E.B.E. Limited, Postbus 6275, Rotterdam, Holland.

ISSUES WANTED

...issues of P.W. (about 1955) dealing with mods to the R1155.—D. Sager, 92 Marnham Crescent, Greenford, Middx.

...P.W. for Jan. and Feb. 1971 and P.E. for Jan. to Oct. 1970 inclusive and P.E. for Jan. and Feb. 1989.—J. R. Rourke, 12 Daisy Grove, Edge Hill, Liverpoot, I.7.10R.

...buy or loan the June 1956 issue of P.W. containing the article on PCR Mods.—T. Roche, Arno, Florence Road, Bray, Co. Wicklow, Ireland.

DIGITAL FREQUENCY COUNTER TIMER

continued from page 411

With Q of IClb high, the signal gate IC2a allows the input signal to produce an output which passes to the counter.

The Q output from IC1b is now low and this clears or resets ICIa directly so that the Q output of ICIa now becomes low and the  $\overline{Q}$  becomes high.

As the Q of IC1b is low, IC2b and IC2c maintain

the clamping action of D4.

The condition of ICla with Q low presets IClb so that on the next negative going edge of the clock period signal, IC1b will set to Q low, thus producing a continuous high output from the signal gate IC2a and terminating the gate period by isolating the input signal from the counter.

The  $\bar{Q}$  of IC1a and the  $\bar{Q}$  of IC1b are now both high causing IC2b output to be low and IC2c output to be high and rendering the clamping diode D4 non-conducting. The waveforms of the control oscillator are shown in Fig. 8. The state of the control oscillator whilst in the clamped condition is with Tr6 non-conducting and Tr7 conducting. When D4 is made non-conducting, the current through R13 and D3 causes Tr6 to conduct producing a negative-going step at Tr6 collector from +5V to 0V.

TO BE CONTINUED



## High fidelity Monolithic Integrated Circuit Amplifier

Two years ago Sinclair Radionics announced the World's first monolithic integrated circuit Hi-Fi amplifier, the IC.10. Now we are delighted to be able to introduce its successor, the Super IC.12. This 22 transistor unit has all the virtues of the original IC.10 plus the following advantages:

- 1. Higher power,
- 2. Fewer external components.
- 3. Lower quiescent consumption.
- 4. Compatible with Project 60 modules.
- Specially designed built-in heat sink. No other heat sink needed.
- 6. Full output into 3, 4, 5 or 8 ohms.
- Works on any voltage from 6 to 28 volts without adjustment.
- 8. NEW 22 transistor circuit.

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Output power 6 watts RMS continuous (12 watts peak).

Frequency Response 5 Hz to 100KHz  $\pm$  1 dB.

Total Harmonic Distortion Less than 1%. (Typical 0.1%) at all output powers and all frequencies in the audio band.

Load Impedance 3 to 15 ohms.

Power Gain 90dB (1,000,000,000 times) after feedback.

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Size 22 x 45 x 28 mm including pins and heat sink.

Input Impedance 250 Kohms nominal.

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With the addition of only a very few external resistors and capacitors the Super IC.12 makes a complete high fidelity audio amplifier suitable for use with pick-up, F.M. tuner etc. Alternatively, for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super



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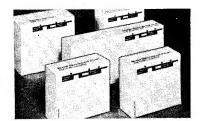
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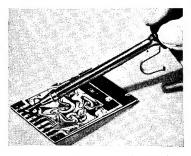


# Sinclair Project 60

## The World's leading range of high fidelity modules

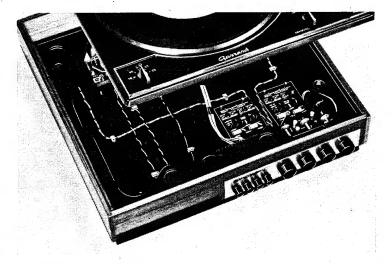






Sinclair Radionics Limited, London Road, St. Ives, Huntingdonshire PE17 4HJ.





Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

Project 60 modules are more versatile – using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all – price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

#### Typical Project 60 applications

System	tem The Units to use together with		Cost of Units
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
20 + 20 W. stereo amplifier for most needs	2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
20 + 20 W. stereo amplifier with high, performance spkrs.	2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
40 + 40 W. R.M.S. de-luxe stereo amplifier	2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43

F.M. Stereo Tuner (£25) & A.F.U. Filter Unit (£5.98) may be added as required.

## from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules

#### Z.30 & Z.50 power amplifiers



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at full output and all lower outputs. Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

## SPECIFICATIONS (Z.50 units are inter-changeable with Z.30s in all applications). Power Outputs

Z.30 15 watts R.M.S. into 8 ohms using 35 volts; 20 watts R.M.S. into 3 ohms using 30 volts.
Z.50 40 watts R.M.S. into 3 ohms using 40 volts;

30 watts R.M.S. into 8 ohms using 50 volts. Frequency response: 30 to 300,000Hz±1dB. Distortion: 0.02% into 8 ohms.

Signal to noise ratio; better than 70dB unweighted.
Input sensitivity: 250mV into 100 Kohms.
For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57 mm.

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#### **Power** Supply Units

Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ.5 30 volts unstabilised £4.98 PZ.6 35 volts stabilised £7.98

PZ.8 45 volts stabilised (less mains transformer) £7.98 PZ.8 mains transformer £5.98

# PZ.5

#### The Sinclair Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a smell charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.

#### Project 60 Stereo F.M. Tuner



First in the world to use the phase lock loop principle

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Good reception is possible in difficult areas, and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.

SPECIFICATIONS—Number of transistors: 18 plus 20 in LC. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity:  $2\mu V$  for 30dB quieting:  $7\mu V$  for full limiting. Squelch level:  $20\mu V$ . A.F. C. range:  $\pm 200$  KHz. Signal to noise ratio: > 65dB. Audio frequency response: 10 Hz. -15 KHz ( $\pm 148$ ). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level:  $2\mu V$ . Cross talk: 40dB. Output voltage:  $2\times 150$ mV R.M.S. Operating voltage:  $25 \times 30$  VDC. Indicators: Mains on; Stereo on; tuning. Size:  $93 \times 40 \times 207$  mm.

Built and tested. Post free.

£25

### Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

means of push buttons and accurate equalisation is provided to an interessed injuries. 
PECIFICATIONS—Injurt sensitivities: Radio—up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A curve ±1dB:20 to 25:000 Hz. Ceramic p.u.—up to 3mV: Alix—up to 3mV. Output: 250mV. Signal to noise ratio: better than 7ddB. Channel matching: within 1dB. Tone controls: TREBLE + 15 to —15dB at 10 KHz: BASS ∓ 15 to —15dB at 100Hz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm. 

19.98

Built tested and guaranteed.

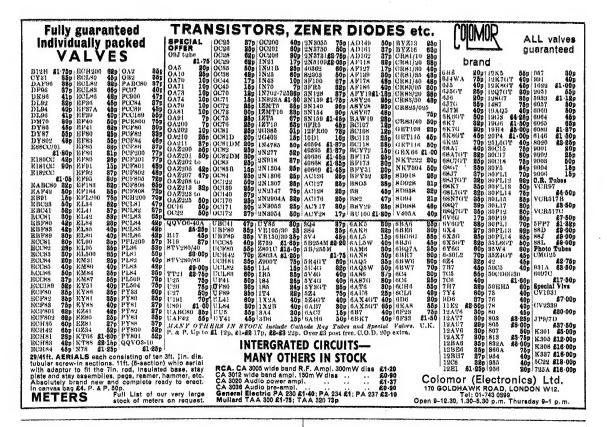
## A.F.U. High & Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase dis-

of the walled signal man has previously been possible. An include a phase tortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current – 3mA. H.F. cut-off (–3dB) variable from 28Hz to 5KHz. L.F. cut-off (–3dB) variable from 25Hz to 100Hz. Distortion at 1KHz (35V. supply (0.02% at rated **F.5.98**) Built tested and quaranteed. output. Size: 66 x 40 x 90 mm.

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17 transistors, 8 diodes

PERFORMANCE CAN BE COMPARED WITH \$65 RADIOS:

PERFORMANCE CAN BE COMPARED WITH 265 RADIOS: Sockets for tape recorder, record player, extension speaker, earphone, external power supply (128 y DO), external line and dipole aerials. Battery life approx, 300 hours.

At last, the Astrad Solar Radio combining quality with value and a specification not before available from Russia. Designed for the man who appreciates the extras necessary for quality performance. This 6 waveband Radio has twin speakers for quality. Fully independent Treble and Bass tone controls for All & FM tondensts. 12 push buttons for waveband selection including automatic frequency control on VIII. This is the big set from Russia, overall dimensions—14° x 3½ x 4½. Included in this incredible specific that the selection including motion to ensure that you are described in a minufic thing motion to ensure that you are described in a minufic thing motion to ensure that you are described in a push-button dial light. All proceeded in the push available at a push-button dial light. All proceeded in the push available at a neverto be repeated price. Immediate delivery.

£23-97 Earphones 25p. P. & P. 75p

SEBA ELECTRONICS, Dept. M/O, 104 Uxbridge Road, Hanwell, London, W7.
Phone: 01-579 5644

## REE FLOG LIST No.6

Scores Special

SEND S.A.E. OR CALL AT ANY BRANCH FOR YOURS

HUGE RANGE OF SPECIAL SNIPS— STEREOS, DECKS, TAPE RE-CORDERS, AMPLIFIERS, SPEAK-ERS, TRANSISTORS, VALCE COMPONENTS—AT ALL BRAN-CHES. ECHNIGAL *TRADING* 

ortsmouth-350 Fratton Rd. Tel. 22034

- ★ Southampton-72 East St. Tel:25851 Southampton-77 East St. Tel. 31276
- Brighton-6 Queens Rd. Tel: 23975
  ★ All Mail Order Tel: 6903 882299
  15-17 QUEEN ST. ARUNDEL

ENORMOUS PURCHASE, GUARAN-TEED. APPROX. HALF PRICE. WORLD \* TAPE \* FAMOUS

Standard Play

Sin. 150ft. 12p

Sin. 200ft. 22p

Sin. 300ft. 22p

Sin. 1,200ft. 52p

Sin. 1,200ft. 52p

Sin. 1,200ft. 52p

Sin. 2,400ft. 21-35

Triple Play

Sin. 225ft. 14p

Sin. 200ft. 28p

Sin. 1,800ft. 28p

Sin. 1,800ft. 28p

Sin. 1,800ft. 28p

Sin. 1,800ft. 52p

Sin. 1,300ft. 65p

Sin. 1,200ft. 65p

Sin. 1,300ft. 93p

Quadruple Play

Postage 5p reel. 3in. 600ft. 48p Post Free less 5% on three reels. Quantity and Trade enquiries invited

15

#### THIS MONTH'S NEW BARGAINS

THIS MONTH'S NEW BARGAINS TO 3 Heat Sink. Suitable for most power transistors OC 26 etc. This is aluminium, anodised black for maximum heat dissipation. Supplied complete with mice disc and insulation bushes. Price 109 each or 10 for 90. Size approx. 22 \*\* 22 \*\*. 12 Volt 500 M.A. Mains Transformers. Ministure type now available, price 559 each or 10 for 25. 3 Core Flex, New Golouts. (Brown for live, blue for neutral, yellow/green for earth). This is completely P. V.C. covered and ribbed, virtually non-kinkable Suttable for washing maximum oversets. Conductor size 23/36. £1 per dozen yards or £10 per 100 yards coil.

Valve Holders at remarkably low prices—moulded construction—made by most famous company. Price per each

		1-9	10-99	100-999	1,000 up
B7G	Flanged	2p	1.5p	1.25p	1p
B7G	Skirted	3p	2p	1.75p	1.5p
B7G	Printed				
	Circuit	2p	1.8p	1-4p	1.2p
B9A	Flanged	20	1-68	1.3p	1-1p
B9A	Skirted	3p	2.5n	2.25p	2p
B9A	Printed				
	Circuit	20	1.9p	1.60	1-3p

MOTOR by A.E.I. 1/20 bp, 1.8p 1.8p 1.8p

MOTOR by A.E.I. 1/20 bp, 1275 rpm Self starting for normal A.C. Mains, a well made enclosed motor with standard A. d. Mains, a well made enclosed motor with standard A. d. Mains, a start 1 long, Suitable for light power operation. Continuous rating, 21.50 pius 20p post.

Instrument Motor by Evershed Hysterals motor or makers the motor for 110 v A.C working double ended shaft—23.50 pius 20p post and insurance. Cut and Prepared 3 Gore Leads, 2 yds. long, P.V.C. covered and ribbed virtually non-kinkable 23/36 conductors. Old colour scheme. Sp each, 10 for 70p. New colour scheme 15p each or 10 for 21.455.

Hearing Aid Amplifiers. Ext behind ear deat Aids.) 3 transistors on tiny P.C. board with volume control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as one control—whole thing only about half as big as an extended the control of the co

for £1.08. 24-Way Rotary Switch—Single pole £1.25—double pole £1.95. Double Pole, Double Throw Toggle Switch. Suitable for mains voltage and up to 10 amps. 15p or 10 for £1.35.

BREAK GLASS FIRE ALARM
PUSH
Made by AFA. As used all over the
country. Made from heavy cast steel.
Drop front. Open with Allen key for
test. Switch normally closed, openswhen glass is broken. Diameter approx.
5° £1.25 or with cast steel mounting
box £1.75, Post 20p.



## CAR ELECTRIC PLUG Fits in place of cigarette lighter. Useful method for making a quick connection into the car electrical system. 38p each or 10 for 23-42.

ROCKER SWITCH





## PUSH BUTTON CHANGE OVER SWITCHES

This is a Honeywell micro switch mounted on a metal frame with spring loaded plunger to operate. Panel hring by single 2\* hole. Single Changeover switch 255 each or ten for \$2.52. 2-changeover switch operated of the control of the control of the switch of the control of the switch of the control of the switch of the control of the \$2.34.5. 3-changeover switches 45p each or ten for \$4.05.



#### HORSTMANN "TIME AND SET" SWITCH

(A 15 amp Switch). Just the thing if you want to come home to a warm house without it costing you a fortune. You can delay the switch on time of your electric fires, etc., up to 14 hours from setting time or you can use the switch to give a boost period of up to 3 hours. Equally suitable to control processing. Regular price probably around £5. Special snip price £1.50, p. & ins. 23p.

#### OUT OF SEASON BARGAIN TANGENTIAL HEATERS



TANGENTIAL HEATERS

Once again we are able to make a special bargain offer of these very popular heating units. Tangential heaters although brought out a few years ago are still the latest and best type as nothing has yet been made which could be called an improvement on them. The Tangential unit is still the only one used in good distingtion of the tangential unit is still the only one used in good of the tangential unit is still the only one used in good with the tangential unit is still the only one used in good with the tangential unit is still the only one used in good with the tangential unit is still the only one used in good with the tangential unit is still the only one used in good with the tangential unit of the tangential unit is still the only one used in good with special bear within a switching and read element which allows switching and rated only the simplest of cases, even a wooden cabinet. Lot all the suitable or the plints of the kitchen or subject to the plints of the kitchen cabinet. Lot of customers missed our special summer 2001/240 3 k.w. model 35.50. Control switch heaters only 25p or two-heat, cold-blow and off 85p. Postage and insurance 33p on heaters.

#### AMPLIFIER MAINS TRANSFORMERS



50v 1½ amp. Upright mounting with fixing brackets and metal shrouds to contain magnetic field, 50 of s primary, tapped 110v, 117v, 210v, 230v and 250v. 2 secondaries, one 50v 1½ amp, other 6v 1 amp for pilot light etc. £1.95, postage 30v.

#### THIS MONTH'S SNIP LIGHT DIMMER



For any lamp up to 200 watt. Mounted on switch plate to fit in place of standard switch. Virtually no radio interferences. Price 21.99 plus 20p post & insurance.

#### CAPACITOR DISCHARGE CAR IGNITION



This system which has proved to be amazingly efficient and reliable was first described in the Wireless World about a year ago. We can supply kit of parts for improved and even more efficient version, price 2498. When ordering please state whether for positive or negative systems. Plus 30p post.

#### STANDARD WAFER SWITCHES-

Standard size 11 wafer—silver-plated 5-amp contact standard 1" spindle 2" long—with locking washer and nut way 3 way 4 way 5 way 6 way 8 way 9 way 10 way 12 way No. of Poles 3 way 4 way 5 way 6 way 8 way 9 way 10 way 12 way
40p 40p 40p 40p 40p 40p 40p 40p 70p
40p 40p 40p 40p 70p 70p 70p 70p
40p 40p 70p 70p 70p 70p 70p 81p 81p 81p
70p 70p 70p 85p 85p 85p 85p 81p
70p 70p 70p 85p 81p 81p 81p 81p 81p 81p
70p 95p 81p 81p 81p 81p 81p 81p 82p 82p
70p 95p 81p 81p 81p 81p 81p 81p 82p 82p
95p 85p 85p 81p 81p 81p 81p 81p 82p 82p
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95p 85p 81p 81p 81p 81p 81p 81p 81p 82p 82p 82p
95p 85p 81p 81p 81p 81p 81p 81p 81p 81p 82p 82p 82p 1 pole
1 pole
2 poles
3 poles
4 poles
5 poles
6 poles
7 poles
8 poles
9 poles
10 poles
11 poles

#### INSTRUMENT SWITCHES

Precision made with dieast indexing mechanism. Full length 4° spindle 5 amp and silver plated contacts. Range except for 9 way is a standard wafer switches. Prices obviously higher. For 40p read 60p, for 70p read 21, for 95p read 61:40, for 61:45 read 82:20. Note also 2 way types available up to 36 pole, 3 way 30 pole, 4 way 24 pole, 5 way 19 pole, but 10 and 12 way only available up to 6 pole.

#### 3 STAGE PERMEABILITY TUNER



J SIAGE FERMEABILIII I UNEX
This Tuner is a precision instrument made for the famous
Radiomobile Car Badio. It is a medium wave tuner (but set
of longwave coils available as an extra (if required) with a
frequency coverage 1620 Kojs-525 Kojs and intended to
operate with an LF, value of 470 Kojs. Extremely compact
(size only 23 × 2 × \(\frac{1}{2}\) ins. thick) with reduction gear for
fine tuning. 65p, with circuit of front end suitable for car
radio or as a general purpose tuner for use with Amplifier.



ELECTRIC CLOCK WITH 25 AMP SWITCH
Made by Smith's, these units are as fitted to many top quality
cookers to control the oven. The clock is mains driven and
frequency controlled so it is extremely accurate. The two small
disle enable switch on and off times to be accurately set. Ideal
for switching on tape recorders. Offered at only a fraction of the
recular price—new and nunsed only 28.50, less than the value
of the clock alone—post and insurance 13p.

#### DISTRIBUTION PANELS



Just what you need for work bench or lab.  $4\times 13$  amp sockets in metal box to take standard 13 amp insed plugs and onfoir switch with neon warning light. Suppled standard 13 amp insed plugs and onfoir switch with neon warning light. Suppled complete with 7 feet of heavy cable. Wired up ready to work, £2 less plug 22-95 with fitted 13 amp plug, 22-40 with fitted 15 amp plug, 29-40 with fitted 15 with fi

#### MAINS OPERATED SOLENOIDS



Model 772—small but powerful 1" pull—approx size 1\frac{1}{2}" x
1\frac{1}{2}" x 1\frac{1}{2}" 60p.
Model 400/1 \frac{1}{2}" pull. Size
2\frac{1}{2}" x 2" x 1\frac{1}{2}" 75p.
Model TT10 1\frac{1}{2}" pull Size
2\frac{1}{2}" x 2" x 2\frac{1}{2}" x 2\frac{1}{2}" -\frac{2}{2}1.80
plus 20p post and Insurance.

# 5

#### TREASURE TRACER

Complete Kit (except wood battens) to make the metal detector as described in last month's issue, £2.50 plus 20p

## Mains Connector



discon

#### MINIATURE



MINIATURE
WAFER SWITCHES
2 pole, 2 way—4 pole, 2 way—3 pole, 3 way—4 pole, 3 way—2 pole, 4 way—3 pole, 4 way—3 pole 6 way—1 pole, 12 way, All at Be each, £1-80 dozen, your assortment.

#### WATERPROOF HEATING ELEMENT

26 yards length 70W. Self-regulating temperature control. 50p post free.

#### COMPUTER TAPES



50

COMPUTER TAPES
2,400° of the Best Magnetic Tapmoney can buy—users claim good result with Video and sound. 1' wide 21.45 plus 30p post and insurance, with ossettle. ½' wide 21.9 plus 30p post and insurance with cassettle. ½' wide 21 plus 20p post and insurance with cassettle. ½' wide 21 plus 20p post and insurance with cassettle. Spars spools and cassettes—1' 21, 3'' 85p. ½'' 75p each plus 20p post and insurance.

#### BALANCED ARMATURE UNITS

These Capsules are 1% in diameter and 1 thick They will operate as a microphone or loud speaker so can be used in intercom and similar circuits. 30 pten for \$3.

#### MULTI-SPEED MOTOR



MULTI-SPEED MOTOR
Replacement in many well-known
food mixers. Six speeds are available 500, 580 and 1,100 r.p.m. from
either or both of the nylon sockets
(where the beaters of the food
mixers normally go) and 8,000,
12,000 & 15,600 r.p.m. (ideal
polishing speeds) from the main
drive shaft. This drive shaft is \$i\$ in.
diameter and approximately 1 in.
diameter and approximately 1 in.
motor is that
to sing 230/240v. AC-DG ceries wound its speed may
be further controlled with the use of our Thyrister
controller. This is a very powerful and useful motor
size approx. 2 in. dia. X 5 in. long, mains 230/240v.
Price 88p plus 23p postage and insurance. 12 or
more post free.

#### MAINS OPERATED CONTACTOR





#### QUICK CUPPA





#### DIGITAL COUNTER TIMER

as described in this issue. Send today for price list of parts.

Where postage is not stated then orders over £5 are post free. Below £5 add 20p. Semi-conductors add 5p post. Over £1 post free. S.A.E. with enquiries please.

## J. BULL (ELECTRICAL) LTD. Dept. P.W. 7, Park Street, Croydon CRO IYD

# WAREHOU



Rec	Retail	Comet
STEREO AMPLIFIERS ALBA UA 700 ARMSTRONG 521 ARMSTRONG 521 DULCI 2870 DULCI 2870 GOODMANS Maxamp LEAK Stereo 30 Plus LEAK Stereo 30 Plus LEAK Stereo 30 Plus LEAK Stereo 30 Plus LEAK Stereo 70 PLEAK STEREO 70 LEAK STEREO 70 LEAK STEREO 70 LEAK STEREO 70 PLEAK STEREO 70 LEAK STEREO 70 PLEAK STEREO 70 PLEAK STEREO 70 PLINER STEREO 70 PHILLIPS RH 580 PHILLIPS RATOD PLONEER SATOD PLONEER SATOD PLONEER SATOD PLONEER SATOD STOCKER REVENSION ROGERS REVENSION ROGERS REVENSION SINCLAIR PROJECT 60/2 x Z50/PZ5 SINCLAIR PROJECT 60/2 x Z50/PZ5 SINCLAIR AFU SINCLAIR BOOT TELETON GA 701  TELETON GA 701  TELETON GA 701  TELETON GA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON BA 702  TELETON SA 92 206 (new release) TELETON FA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON SA 92 206 (new release) TELETON FA 701  TELETON F	. Retail Price	Price
ALBA UA 700	34·50 56·00	26 · 00 42 · 50
*DULCI 207	26·00 32·00	17-50 21-95
FERROGRAPH F307	62:00	44 · 00
LEAK Stereo 30 Plus	56-50	43 00
LEAK Stereo 70	69.00	52 00
*LINEAR LT 66	75·00 21·00	56·95 17·00
METROSOUND STOOF	36.00	24-95
PHILIPS RH 591	79.00	61 - 50
PHILIPS RH 580	29.00	23.00
PIONEER SA500	62·10 98·00	41 · 95 68 · 95
PIONEER SA900	134 - 10	95.95
ROGERS Ravensbourne	59 - 50	45 95
ROGERS Ravensbrook Mk. II	47.50	36 - 50
ROGERS Ravensbrook (cased)	52 - 50	41 - 50
SINCLAIR 2000	35.00	27 - 00
SINCLAIR PROJECT 60/2 x Z50/	24.00	00.05
SINCLAIR AFU	5.95	4 95
SINCLAIR Neoteric	61 · 95 45 · 00	46·00 34·95
TELETON SAQ 206 (new release)	29.00	18.50
VOLTEX 100w. Stereo Discotheque,	405.00	420.00
Starred items above take ceramic	cartridge	s only.
All others take both ceramic and ridges.	magnet	ic cart-
TUNERS		
*ARMSTRONG 523 AM/FM	53 - 76	42.00
ARMSTRONG M8 Decoder	9.50	8.00
DULCI FMT.75 Stereo	35-00	22 · 00 29 · 25
GOODMANS Stereomax LEAK Stereofetic Chassis	82·52 66·50	49·95 52·00
LEAK Stereofetic in teak case	72 - 50	59·00
PHILIPS RH 691	89.00	75.50
PIONEER TX900 AM/FM	153 69	123 00
ROGERS Ravensbourne chassis ROGERS Ravensbrook	61 · 89 45 · 01	47·95 38·00
ROGERS Ravensbrook (cased) TELETON 201X FM	51 · 26 36 · 00	41 · 00 27 · 50
TELETON GT 101	45.50 th MPX	31 · 95 Stereo
ridges.  TUNERS *ARMSTRONG 523 AM/FM *ARMSTRONG 524 FM ARMSTRONG MS Decoder *DULCI FMT.7 FM DULCI FMT.7 FM DULCI FMT.7 Stereo GOODMANS Stereomax LEAK Stereofetic Chassis LEAK Stereofetic In teak case PHILIPS RH 590 PHILIPS RT 590 PHILIPS RT 590 PHILIPS RAYON AM/FM ROGERS Ravensbrook ROGERS Ravensbrook (cased) TELETON 201X FM TELETON GT 101 All above Tuners are complete wi Decoder except where st  TUNER/AMPLIFIERS AKAI AA 5500	arred.	
TUNER AMPLIFIERS AKAI AA 8500 AKAI A6 6500 ARENA 8500 ARENA 2400 ARENA 2400 ARENA 2700 ARENA 2700 ARENA 79000 ARENA 7900 BOOLORE SAMPLE AND AMPLE AND AMPL	229.00	181 - 00
AKAI 6600	142 - 53	112 - 00
ARENA 2400	90.30	72.00
ARENA 2700	229·00 142·53 82·00 90·30 105·00 303·45 9·50 91·89 104·71	181 · 00 112 · 00 67 · 00 72 · 00 85 · 00 258 · 00
ARMSTRONG M8 Decoder	9·50 91·89	8·00 74·50
ARMSTRONG 526	104 - 71	84 - 50
MIDLAND 19/542	104 · 71 95 · 00 49 · 56 139 · 00 160 · 43 194 · 74 111 · 10	258 00 8 00 74 50 84 50 72 00 37 50 112 00 126 00 150 00 89 00
PIONEER SX770 AM/FM	160 - 43	112 00 126 00
PIONEER SX990 AM/FM	194 · 74	150·00 89·00
TANDBERG 1171 comp. with	110.00	94.00
*TELETON F2000	51.50	30 - 50
TELETON 10AT1 150w. RMS	160 00	103.00
TANDBERG 1171 comp. with decoder *TELETON F2000 TELETON 7AT20 TELETON 10AT1 150w. RMS TELETON TFS50 TELETON TFS50 TELETON TFS50 TELETON TFS 50 LA MW/LW/VHF*TELETON CR55 WHARFEDALE 100.1 Starred items above take ceramic All others take both ceramic rake in the ceramic and the ceramic rake ce	82·50 87·50	94·00 30·50 77·95 103·00 59·00 63·50 37·50
*TELETON R.8000 with Speakers TELETON CR55	63 - 25	37 50 87 00
WHARFEDALE 100.1	149.00	119.00
All others take both ceramic and	magnet	ic cart-
ridges. All the above Tuners and Tuner/Ar	nplifiers	include
ridges. All the above Tuners and Tuner/Ar MPX Stereo Decoder with the ex- strong where decoder is extra as I	eption o	of Arm-
PICKUP ARMS GOLDRING Lenco L75 GOLDRING Lenco L69 SME 3009 with S2 Shell SME 3012 with S2 Shell	12·33 9·29 34·47 36·71	10·50 7·00 27·00 30·50
SME 3009 with S2 Shell	34 47	27.00
SME 3012 with S2 Shell	36.71	39 · 50



Rec.	Retail	Comet	
CARTRIDGES	Price	Price	
GOLDRING 800	13.00	6.95	
	18.86	10.95	
GOLDRING 800E	26.01	19-50	
*GOLDRING CS90 Stereo	5.20	4 - 25	
*GOLDRING CS91/E	7.81	6 25	
GOLDRING G850	6.50	5 25	
EMPIRE 1000ZE/X	63.00	52 50	
EMPIRE 999VE/X	44 - 50	36.00	
EMPIRE 999TE/X	27 60	22 - 50	
EMPIRE 999SE/X	21.00	17.50	
EMPIRE 999E/X	16 50	13.00	
EMPIRE 999/X	11.50	9.25	
EMPIRE 909E/X	12 85	10.25	
EMPIRE 909/X	9.00	7.50	
EMPIRE 90EE/X	9.75	8-00	
	cial Pri		
PICKERING V15 AC2	8 · 40	7.00	
ORTOFON SL15E	29.65	23 · 75	
ORTOFON 2X15K Transformer	7.00	5 · 25	
SHURE M3DM	7 - 41	4-95	
SHURE M31E	12 · 05	8 · 95	
SHURE M32E	11 · 10	8 · 25	
SHURE M32-3	10.20	8.00	
SHURE M44-5	11 · 10	7 95	
SHURE M44-7	10.20	8.00	
SHURE M-44C	10.20	8.00	
SHURE M44E	12.05	8 · 25	
SHURE M55E	12.95	9.50	
SHURE M75G	17 · 60 16 · 70	14·00 13·00	
	19.45	16.00	
SHURE M75EJSHURE M75E-95G	23-15	18:00	
	21.30	14 50	
SHURE M75ESHURE V15-11	40.76	28 95	
SHURE V15-11-7	38-90	29 00	
Starred cartridges above are ceramic			
magnetic.		u.c	
/			

#### TURNTABLES

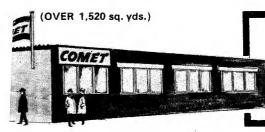
GARRARD AP76 turntable, fully wiredcomplete with Goldring G.800 Cartridgebase and cover. Special Price 30-95
GARRARD SP25, Mk III fully wired with
Goldring G800 Magnetic Cartridge. Complete
with base, plinth and cover—
Special Price 21-98
GARRARD 2025 fully wired with Sonotone
9TAHC Cartridge complete with base and
cover
Special Price 15-95
GOLDRING 705/P turntable, fully wired complete with Goldring G850 Cartridge, base and
cover. List Price £25-00 Special Price £15-95

DUAL 1219 transcription	60 · 40	50.00
DUAL 1209 transcription	42 - 62	35 - 00
GARRARD SP25 Mk III	16-45	11 - 90
GARRARD SL65 B	21.25	15.50
GARRARD SL75 B	39.20	26 - 50
GARRARD SL95 B	50.01	35 50
CARRARD SESS B		
GARRARD 401	38.07	29 · 25
GARRARD SL72 B	33-11	25 - 50
GARRARD 3500, with GKS Cart.	17-23	12-90
Base and Cover to fit GARRARD		
SP25, SL55, SL65B and 3500 Sp	ecial Pri	ce 4 · 00
GARRARD 40B	13.84	10.97
GARRARD AP76	28 - 88	21 - 50
GOLDRING GL69 Mk. II	26-63	21.25
GOLDRING GL69P Mk. II		28 - 50
GOLDRING GLOSP MK.II	35 14	
GOLDRING GL75	36 - 41	31 · 50
GOLDRING GL75P	46.94	38 · 50
GOLDRING Covers for 69P and 75P	4 · 21	3.50
GOLDRING C99—plinth and cover		
for G99	11-45	9 90
GOLDRING G99	26.00	22.90
GOODMANS 3025	37.74	26.90
McDONALD MP 60	15.75	11 . 75
MICDONALD MIC OU		
McDONALD 610	20.00	15.75

#### COMET for after-sales service THROUGHOUT THE U.K.

Pictured, Service Dept. at Clough Rd., Hull also at Leeds, Stockton, Goole, Wakefield, Doncaster, and Bridlington

All the same	and Bridin	9:01.	
	Rec	. Retail Price	Comet Price
Base and Cover f MP 60 and 610 PHILIPS 228 PHILIPS GA 146 PHILIPS 202 Elect PIONEER PL 12A THORENS TO 125 THORENS TO 125A THORENS TO 125A THORENS TO 150A THORENS TO 150A THORENS TO 150A THORENS TO 150A	or McDONALD	File	FIICE
MP 60 and 610	Spe	cial Pri	ce 4 · 50
PHILIPS 228		20.00	17.00
PHILIPS 902 Flort	ronic	80.00	54.00
PIONEER PL12A w	ith base & cover	49 93	37 95
THORENS TX 25	over	8.85	7.00
THORENS TD125		79.04	61 95
THORENS TO125A	Mir II	120 · 20	98.00
THORENS TOISON	R Mk II	40.10	42.05
THORENS TX11 Co	ver	4.43	3.95
			- ••
SPEAKERS			
SPEAKERS B & W Model 70 B & W DM3 B & W DM1 CELESTION Dittor CELESTION Dittor GOODMANS Maig		159 - 50	121 .95
B & W DM3		63.00	49 - 95
CELECTION Ditto	100	32 · 00 24 · 00	25·50 18·00
CELESTION DISTOR	15	32.00	25 . 50
CELESTION Dittor	25	65.00	50.95
GOODMANS Min	ster	25.00	20.50
GOODMANS Magi	ster	62 - 50	47 95
GOODMANS Max	im	20.39	16 - 50
GOODMANS Max GOODMANS Mezz GOODMANS Magi	0 3	34·00 44·00	25 · 95 33 · 00
KEF Cresta	tuili NZ	22 - 17	18.00
KEF Cresta KELETRON KN500	4 speaker		
system (pair) KELETRON KN700	4 speaker	23.00	15·50 18·25
system (pair) KELETRON KN 1000	4 speaker	27 · 50	
system KELETRON KN1500	4 speaker	19 - 25	13.95
system		24 · 75	18-95
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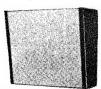


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2N2904A	42p	4 874 4808	TOP	1 20141	200				

Values available

E12 E24 Kie

)	EQ	RTO	PQ	10%	_5%	- 2%

Code	Power	Tolerance	Range
C	1/20W	5%	82 Ω-220K Ω
00000	1/8W	5%	4·7 Ω-470K Ω
Č	1/4W	10%	4-7 Ω-10M Ω
Č	1/2W	5%	4·7 Ω-10M Ω
Č	1W	10%	4-7 Ω-10M Ω
MO	1/2W	2%	10 Ω-1M Ω
WW.	1W	$10\% \pm 1/20 \Omega$	0.22 Ω-3.9 Ω
WW	3W	5%	12 Ω-10Κ Ω
WW	7W	8%	12 Ω~10K Ω
Codes		o film blab stabil	ity low noise

Codes: C == carbon film high stability low noise

MO == metal oxide Electrosii TR5 ultra low noise

WW == wire wound Plessey.

Values: E12 denotes series: 10, 12, 16, 18, 23, 27, 33, 38, 47, 56. 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

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22 NOR DECEMBER 24 Values; 400mW: 2.7V to 30V 15p each 1W: 6.8V to 32V 27p each 1.5W: 4.7V to 75V 60p each Clip to increase 1.5W rating to 3 watts (type 266F) 4p.

#### CARBON TRACK

POTENTIOMETERS, long spindles

Double wiper ensures minimum noise level.
Single gang linear 220 G, to 2-2M G
Single gang log 4-7K G to 2-2M G
Dual gang log 4-7K G to 2-2M G
Dual gang log 4-7K G to 2-2M G
Dual gang log 4-7K G to 2-2M G
Log/antilog 4-7K G to 2-2M G
Log/antilog 4-7K G to 2-2M G
Any type with 1A D.F. mains switch, extra

Please note: only decades of 10. 22 and 47 are available within ranges quoted.

#### CARBON SKELETON PRE-SETS

Small high quality, type PB, linear only  $100\Omega$ ,  $220\Omega$  470 $\Omega$ , 1K, 2K2, 4K7, 10K, 22K, 47K, 100K, 220K, 470K, 1M, 2M2, 5M,  $10M\Omega$  Vertical or horizontal mounting 5p

ENAMELLED COPPER WIRE even No S.W.G. only. 20zreels 16.22 S.W.G. 25p; 24-30 S.W.G. 30p; 32-34 S.W.G. 33p; 36-40 S.W.G. 35p; 40z reels 16-22 S.W.G. only 41p.

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NEWMARKET TRANSISTORS RADIOHM POTENTIOMETERS

#### SAVE £6.25 ON THE NEW ENGLEFIELD 840 AMPLIFIER

This latest Peak Sound high fidelity amplifier incorporates many strikingly original features including facility for adding in stereo FM tuner. Superb performance characteristics and luxurious styling Electrovalue choice of simulated soft leather covering in black, blue, red or dark green. As advertised at \$45. Our price.

£38.75 nett

## MAINLINE AMPLIFIER KITS 70W. £19:60 nett. (main amplifier module).

#### 30 WATT BAILEY AMPLIFIER PARTS

Transistors and PCB for one channel \$6.46 Capacitors, resistors (metal oxide), and transistors, one channel. \$8.41.

Complete unregulated power supply pack, \$4.75. Suitable heatsink 10DN 400C. 55p.

#### INTEGRATED CIRCUITS SIEMENS' TIL INTEGRATED

CHECUITS. Full range of all types in stock for very wide range of applications. For full details, see our latest components list. Please seed S.A.E. PLESSEY SI462A 7.5 ohms. SE-10 nett. Applications data 10p.

## S-DeCs put an end to birds nesting

Components just plug in—saves time—allows re-use of components. S-Dec (70 points), \$1.00
T-Dec, may be temperature-cycled (208 points), \$2.50.
Also µ-Dece and IC carriers.

#### INDICATOR LAMPS

NBOR chrome bezel, round red NR/R, 24p; chrome bezel, round amber NR/A, 24p; chrome bezel, round clear NR/C, 24p, Neon, square red vype L850/P, 18p; amber type L850/P, 18p; amber type L850/A, 18p, 41p; clear type L850/A, 18p, 41p; above are for 240V mains operation. Filament ypen 6V, 0-64A square red type L850/A-6V, 20p, 180 C/G-6V, 20p; 3V 0-64A green type L850/R-6V, 20p; 18V 0-64A L850/R-6V, 20p; 18V 0-64A L850/R-28V, 45p.

## MULLARD SUB-MIN ELECTROLYTIC

Prices are in pence each for same chmic value and power rating, NOT mixed values. (Ignore fractions of 1p. on total value of resistor order)

MULLARD polyester C280 series 250V 20% 0-01: 0-022: 0-038, 0-047 8p ea. 0-068; 0-1, 0-15 4p, 0-25 5p, 10-2: 0-038 7p, 0-47 8p, 0-68 11p, 12F 14p.1-04F 81p, 2-24F 24p.

10 to 99 (see note below)

100 up

TULLARU SUB-MIN ELECTROLYTIC
0426 range arial lead
9 sech
9 sech
9 sech
9 sech
9 sech
10 sech

#### LARGE CAPACITORS

1 to 9

1 1·2 2·5 4

High ripple current types: 1000/26 28p; 1000/50 41p; 1000/100 82p; 2000/25 87p; 2000/50 87p; 2000/100 81-44; 2500/64 77p; 2500/76 85p; 5000/20 62p; 5000/50 81-10; 5000/100 82-91; 10000/50 82-40.

#### COMPONENT DISCOUNTS

10% on orders for components for £5 or more. 15% on orders for components for £15 or more (No discount on

#### POSTAGE AND PACKING

Free on orders over £2, Please add 10p if orders under £2. Overseas orders welcome: carriage and insurance charged at cost. Send S.A.E. for latest list. Prices subject to alteration without notice.

## SIEMENS' POLYCARBONATE CAPACITORS

5% tolerance 250V up to 0·1mF. 100V 0·1mF and above. 2001, 0·12, 0·13, 0·022, 0·027 ... ... ... ... ... ... 59 0·23, 0·29, 0·047, 0·056, 0·068, 0·082, 0·1, 0·12, 0·15, 0·18, 0·22 ... ... 79 0·23 ... ... ... 69 59

#### DIN CONNECTORS

plug sacker   2-pole   18p   10p   10p	

CORED SOLDER-64/40 alloy, 20 s.w.g. 8oz reel, 65p, 1 lb reel, \$1.20.

Handbook of transistor equivalent and substitutes, 40p. (Postage 3p if ordered alone).

**ELECTROVALUE** 

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F.A.L. PHASE 50 MK.II AMPLIFIER \$33.50 PR. FANE POP 25/2 25W L/SPEAKERS \$13.50 Terms: Deposit £5.50 and 9 mon bayments of £4.72 (Total £47.98)

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Terms: Deposit £22 and 9 monthly
payments of £11 (Total £121) F.A.L. PHASE 100 AMPLIFIER
4 FANE POP 50 L/SPEAKERS
Terms: Deposit £14 and 9 monthly
payments of £10 (Total £104)

PACKAGE PRICE £44 garr. £47.00 £33.50 £21

PACKAGE PRICE £49.50 garr.

PACKAGE PRICE £110 £1.50 £121.75 £61.95 £42 PACKAGE PRICE

£94 carr. £1.50

#### **AUDIOTRINE HI-FI SPEAKER SYSTEMS**

Consisting of matched 12in. 11.000 line 15 West 15 ohm high quality speaker, cross-over unit and tweeter. Smooth response and extended frequency range ensure £5.75 OR SENIOR 15 WATT INC. £6.75 Carr. 30p HF126 15,000 LINE SPEAKER carr. 35p



#### AUDIOTRINE HIGH FIDELITY LOUDSPEAKERS



Heavy construction. Latest high efficiency ceramic magnets. Treated Cone surround. "D" indicates Tweeter Cone providing extended frequency range up to 15,000 c.p.s. Impedance 3 or 8-15 ohms. Please state choice. Exceptional performance at low cost.

€54.50

£103.95

Exceptional performance at low cost.

HF801D 8° 8W £2.71 HF120D 12° 15W £4.75

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#### **FANE 807 HIGH FIDELITY LOUDSPEAKER**

A full range 8in. 10 wait unit for excellent sound quality, in suitable enclosure, Cast chassis 8:01 P.V.C. cone surround and long throw voice coil to achieve very low fundamental resonance of 80 c.p.a. Tweeter cone is litted to extend high note response. Frequency range 25 Hz to 15 KHz. Impedance 3 cs. 43.50 8-15 \( \text{1.} \), Gauss 10,000. Remarkable value.



#### HI-FI SPEAKER ENCLOSURES

Modern design. Teak veneer finish. Acoustically ined.
All sizes approx. Carr. 30p. per enclosure
128 Size 16 x 11 x 9in. Pressurised. Gives
pleasing results with any Sin. Hi-Fi 'speaker.
Bi-Fi 'speaker, Size 22 x 15 x 9in.
Hi-Fi 'speaker and Tweeter with 12in
with 10in. Hi-Fi 'spir 26.74
Size 25 x 16 x 10i.
Hi-Fi 'speaker and Tweeter.

47.87



for Record Playing units, cut for Garrard 1025, 2025, 3000, AT60, 8P25 etc.

£3-15 £6.30

LEADING MAKES HI-FI EQUIPMENT AT CLEARANCE PRICES AVAILABLE AT BRANCHES ONLY

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An all-dry battery eliminator. Size 5½ x 4½ x 2in. approx. Completely replaces batteries supplying 1.5v. and 90v. where A.C. mains 200/250v. 506/s is available. Complete kit with diagram £3 or assembled ready for use £3.50

## SPECIAL HI-FI **OFFERS**

All carriage paid. Mail orders only.

WHARFEDALE/F.A.L. 1 pr Wharfedale Super Linton Speak-ers, 1 F.A.L. Phase 32 15+15 watt stereo amplifier. Total Rec. Retail Price \$87.70, Package Price \$66.50.

Above brand new goods. Following have been unpacked for display purposes only.

LEAK MINI SANDWICH (Teak finish) \$1.8, ROGERS RAVENSBROOK STEREO AMP. \$29. TANDBERG 62 Tape Recorder (steroo) \$28, GOLDRING GL68 Transcrip-tion Turnishe and P.U. \$14-50, GOODMANS AUDIO M91 \$19.95

#### R.S.C. TA6 6 Watt HI-FI 10 m SOLID STATE AMPLIFIER

200-250v. AC mains operated. Frequency Response 30-20,000 c.p.s. - 24B. Harmonic Distortion 0.3% at 1,000 c.p.s. Separate Bass and Treble 'lift' and 'cut' controls, 3 input sockets for Mike, Gram, Radio or Tape. Input selector switch. Output for 3-15 ohm spkrs. Max sensitivity Faw. Output rating I.H.F.M. Fully enclosed enamelied case, 94 × 24 × 5½m. Attractive brushed sliver finish facia plate 10½ × 3½m. and matching diagrams and instructions.

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#### OR FACTORY BUILT WITH 12 MONTHS' GUARANTEE £9.75 R.S.C. All HI-FI 12-14 WATT AMPLIFIER



PUSH-PULL OUTPUT. Two input sockets with sepvol. controls for mixing. High sensitivity, 5 valves. Bass and treble controls. Response ± 3dB 30-20,000 e/s. Hum level -60dB. Sensitivity 40 millivoits. For Cycled or Ceramic PUs. High Impedance "mikes". For Musical Instruments etc.

Std. AC mains. For 3 & 16 hom spkrs. Complete kit. Full £10-50 instructions and point-to-point wiring diagrams. Carr. 60p 8AE for leaflet.

Twin bandled metal cover £1.76 Factory built £14.75 or Dep. £3 and 9 mthly pymnts of £1.60 (Total £17.40.)

## R.S.C. COLUMN SPEAKERS

IDEAL FOR VOCALISTS AND PUBLIC ADDRESS



All types 15 Ohms covered in Rexine Vynair
TYPE C4100 IS ALSO SUITABLE FOR BASS GUITAR OR ELECTRONIC ORGAN

TYPE C48S 25-30 WATTS

Fitted four 8" high flux 8 watt speakers. Overall size approx 48 x 10 x 5 in. Carr 50p Terms: Dep £3 and 9 monthly £17.75 payments £2 (Total £21)

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Fitted four 12" 11,000 line 15 watt
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## 30 WATT HI-FI AMPLIFIER FOR GUITAR, VOCAL OR INSTRUMENTAL GROUP.

A 2 or 4 input, 2 vol. control Hi-Fi unit with Separate Bass and Treble controls. Current valves. Peak output rating. Strong Rexine covered cabinet with handles. Attractive black/gold P.V.O. tacis. Reon indicator. For 200-250v. A.C. mains. For 3 or 15 ohm speakers. Send S.A.E. for leaflet.

Terms: Deposit £3.70 and 9 monthly payments of £2.10 (Total £22.60)

£19.95 Carr. 65p





HIGH QUALITY LOUDSPEAKER UNITS
IN TEAK VENEERED CABINETS

118 18' × 8' 10 Watt L12 12'
10,000 lines 3 or 15 25 WATT supplied of the supplied of the state inpedance 15 ohms. Texture conservative rating. Impedance 8-15 ohms. £5.25 carr. £10.50 carr. Or deposit 24 and 9 monthly payments of £3.85. Total £34.15



#### R.S.C. BASS REGENT 50 WATT AMPLIFIER

A powerful high quality all-purpose unit for lead, rhythm, basquitar, vocalists, gram, radio, tape, Peak Output rating.

Loudspeaker unit optional horizontal or vertical mounting.

\* Two extra heavy duty 12in. Loudspeakers.

Four Jack inputs and two Volume Controls for simultaneous use of up to four pick-ups or "mikes"
Bass and Treble controls, Send S.A.E. for leaflet.

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FAL PHASE 100 AMPLIFIER 100W Solid State. 4 Separately controlled inputs Plus master volume control. Ind. Bass and Treble Controls. Output for speakerly 3-39 ohms. Protective circuit to guard against damage from accidental shorts. Or deposit \$19. & 9 monthly pay. \$461.95 ments \$625. Total \$88.25 (Carr. 75p Send S.A.E. for leadets.

## FANE ULTRA HIGH POWER LOUDSPEAKERS

All power ratings are R.M.S. continuous. 2 YEARS' GUARANTEE
High flux ceramic magnets. Heavy cast chassis. All carr, free. High flux cerami 'POP' 60

18" 100 Watt 14,000 gauss 8/15Ω

Solid state 4 Separately controlled inputs. Plus master vol. control. Ind. Bass and Treble Controls. Protective circuit to guard against damage from accidental shorts.

£22.05

Dep: £6 and 9 monthly payments £2.10 (Total £24.90) FOR BASS GUITAR, ELECT. ORGAN, ETC.

'POP' 50 15" 60 Watt 12" 50 Watt 13,000 gauss 150 £10.50 14,000 gauss 8/15Ω £12.90

Dep. £3.30 and 9 monthly payments. £1.30 (Total £15).

Dep £2 and 9 monthly payments £1.15 (Total £12.85)
Pair suitable all purposes

FANE LOUDSPEAKERS 'POP' 25/2 12in. 25 WATT

Dual Cone 15  $\Omega$  (for uses other than Bass Guitar or Electronic Organ).

£6.75 Carr. free

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## R.S.C AIO 30 WATT ULTRA LINEAR

K.S.L AIU 3U WAII ULIKA LINEAK
HI-FI AMPLIFIER Highly sensitive. Fush-Pull high
and the property of the proper

#### RSC TRANSFORMERS, L.F. CHOKES & RECTIFIERS

FULLY GUARANTEED. Impregnated and Interleaved where necessary

Primaries 200-250v. 50c/s. Screened MIDGET CLAMPED TYPE 2½×2½×2½in, 250v., 60mA. 6.3v. 2a 950-2020... 60mA 6.3° 2a 950
2011.1 SHROUNED UPRIGHT MOUNTING
250-0-250... 60mA 6.3° 2a, 0-5-6.3° 2a. 21.25
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## Introducing R.S.C. MkIII SUPER 30 HI-FI STEREO AMPLIFIER

COMPLETELY NEW DESIGN FURTHER IMPROVED IN BOTH APPEARANCE and PERFORM-ANCE. REPRESENTING VALUE FAR HIGHER THAN THE PRICES SUGGEST.
Only high grade components by 120 Carr. COMPLETE KIT OF PARTS 125 Carr.

R.S.C. SUPER 30 E ) 00

SATIN SILVER METAL FACIA with black lettering. Black added knobs with bright silver centres.

PUSH-BUTTON SELECTOR SWITCHING
NEON INDICATOR
JACK SOCKET FOR HEADPHONES
CABINETED MODEL VENEERED IN SATIN
TEAK. SUITABLE FOR ANY MODERN PICKUP CARTRIDGE CERAMIC OF MAGNETIC,
WE RECOMMEND USE WITH THE BEST
ANCILLARY EQUIPMENT THAT CAN BE
AFFORDED.

Or FACTORY BUILT with 12 months guarantee
Dep. £5-75 and 9 monthly payments £3-50 (Total £37-25)
07 FACTORY BUILT in cabinet as illustrated
Dep. £6 and 9 monthly payments £3-86 (Total £40-74)

DIED. 25 and 9 monthly payments 23-95 (1014) 240-74)

PRINTED CIRCUITRY, TWENTY SILICON TRANSISTORS.

FOUR DIODES, FOUR RECTIFIERS.

TECHNICAL DETAILS (Applying to each channel where a ppropriate)

CONTROLS: PUSH-BUTTON SELECTOR (1) Disc (2) Radio (3) Tape (4) Mono L

(5) Mono R (6) SPEAKER DIS. (7) Mains on/off.

Bass. Treble, and Balance. Plus Ceramic Mag P.U. Switch.

AFFORDED.

OUTPUT: 15 watts R.M.S. (Continuous) into 8 ohms.

10 watts R.M.S. (Continuous) into 15 ohms.

11 watts R.M.S. (Continuous) into 15 ohms.

12 watts R.M.S. (Continuous) into 15 ohms.

13 watts R.M.S. (Continuous) into 15 ohms.

14 watts R.M.S. (Continuous) into 15 ohms.

15 watts R.M.S. (Continuous) into 15 ohms.

16 watts R.M.S. (Continuous) into 15 ohms.

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29 watts R.M.S. (Continuous) into 15 ohms.

20 watts R.M.S. (Continuous) into 15 ohms.

21 watts R.M.S. (Continuous) into 15 ohms.

21 w

## <u>EREO'PACKAGE'OFFERS</u> Matching as recommended for optimum performance. Send S.A.E for coloured brochure showing other money-saving offers. ★ TA12 AMPLIFIER 6:5+6:5 watt in veneered housing ★ GARRARD SP25 MK III Player unit on Plinth

- Four fully wired units ready to 'plug in.'

  \* SUPER 30 AMPLIFIER (15 + 15
- watt) in veneered housing

  GARRARD SP25 MKIII Turntable on
- Plinth with cover GOLDRING CS90 Ceramic Pickup Car-

GOLDRING CS90 Ceramica Fixed tridge with diamond stylus PAIR OF STANWAY II Speaker Units Special Total Price Carr. £1.50

- ★ Super 30 Amplifier (15 + 15 watt)

Package prices apply providing all individual units are purchased from any branch within 3 months.





★ GOLDRING CS90 Ceramic P.U.
Cartridge with diamond stylus
★ PAIR OF DORCHESTER
Loudspeaker Units
Special Total Price
Carr. £1.25
Or Deposit £7.15 and 9 monthly
payments £6.35 (Total £64.30).
Trans. Plastic Cover £3.15 extra. \*\*Super 30 Amplifier 19 \*\*15 water in veneered housing in veneered housing Stanway II Stanway II Trans. Plastic Cover 25/15 Stanway II ATTRACTIVE TEAK or AFRORMOSIA VENEERED CABINETS and PLINTES AND Pair of Stanway II speaker units Special Total Price 267-5 TERMS AVAILABLE ON ALL PACKAGE OFFERS

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unit on Plinth
GOLDRING CS90 Ceramic P.U.

## 'YORK' HIGH-FIDELITY 3 SPEAKER SYSTEM ★ Moderate size only 25×14×10in. ★ Response 30-20,000 c.p.s. Impedance 15 ohms Carr. 63p

Impedance 15 ohms
Performance comparable with units costing considerably more.
Consists of (1) 12in. 15 watt Bass unit with cast chassis, Roll rubber cone surround for ultra low resonance, and ceramic magnet. (2) 3-way quarter section series cross-over system (5) Appropriate quantity acoustic damping material. (6) Handsome Teak veneered cabinet. (7) Circuit and full instructions. Terms: Dep. 24-50 and 9 monthly payments 22-25 (Total 224-75).

DEMONSTRATIONS AT ALL BRANCHES

RSC G66 6+6 WATT high quality STEREO AMPLIFIER

Individual Ganged controls: Bass, Treble, Volume and Balance. Printed circuit construction employing 10 Transistors plus Diodes. Output rating I.H.F.M. Frequency range 20-20,000 c.p.s. Suitable for Crystal Pick-ups, etc., and for loudspeaker output impedances of 3 to 15 ohms. For standard 200—250v A.C. mains operation. Attractive silver finished

metal facia plate & matching control knobs.

COMPLETE KIT of PARTS INCLUDING
FULLY WIRED PRINTED CIRCUIT

and comprehensive wiring diagram and instructions or FACTORY BUILT IN TEAK £12.99 or dep £2:50 and 9 mthly pymnts. £1:45 (Total £15:55)

## HIGH FIDELITY LOUDSPEAKER UNITS )

Cabinets latest style Satin Teak veneer. Acoustically lined or filled acoustic damping. Ported where appropriate. Credit terms available. DORCHESTER (Illustrated) Size 16 × 11 × 9in. appr. Range 45-15,000

c.p.s. Rating 8-10 watts. Fitted High flux 13×8in.
Dual Cone speaker. Imp. 3 or 15 ohums.
STANWAY II Size 20 × 104 × 94in. approx.
Rating 10 watts. Inc.

313 × 8in. speaker with highly flexible cone surround, long throw voice coil and 10,000 line magnet. High flux tweeter. Handsome Scandinavidesign cabinet. Range 35-20,000 c.p.s. Imp. 15 ohms. Gives £17-85 smooth realistic sound output. See 'package offers' for illustration

SMOOIN realistic sound output. See 'package offers' for illustration 217 33 R.S.C. TA12 MKIII 6-5 + 6-5 WATT STEREO AMPLIFIER FULLY TRANSISTORISED, SOLID STATE CONSTRUCTION HIGH FIDELITY OUTPUT OF 8-6 WATTS PER CHANNEL.

Designed for optimum reformance with any crystal proceedings. Radio tuner, Tape recorder cic. \$3 separate switched input sockets on each channel \$8 separate Bass and Treble coutrols \$\$4\$ Bids Switch for mono use \$\$8 peaker Output 3-15 ohms \$\$For 200-250 V. A.C. mains \$\$For Packer Coutput 3-15 ohms \$\$For 200-250 V. A.C. mains \$\$For 200-250



COMPLETE RIT OF PARTS WITH FOR 18 with a complete RIT of Parts with 12 months guarantee £19.50 or Deposit 23 and £15.50 Carr. 9 mthly pymts £2.15 (Total £22.85). Or in Teak veneer housing £23 Dep. £3 & 9 mthly payments £2.55 (Total £25.95). Send S.A.E. for leaflet

## AUDIOTRINE A55 HIGH QUALITY STEREO SYSTEM

0 0 0 0

5 + 5 WATT OUTPUT GARRARD 5200 CHANGER WITH LOW MASS PICK-UP ARM AND STEREO CARTRIDGE.

CONTROLS: TREBLE, BASS, VOLUME, STEREO, BALANCE.
Operation on 200-250v.
A.C. mains.



Luxurious Teak Veneer Finished Cabinets. Transparent plastic (tinted) cover included for main unit. Silver finished facia plate and matching control knobs.

ARD OF OLIVITY IS CONTROLLED.

Carr. £1:25

Terms: Deposit £5:50 and 9 monthly payments £4:50

(Total £46.) Output rating. I.H.F.M finished facia plate and matching control knobs. (Total £46.)

A REALLY SURPRISING STANDARD OF QUALITY IS OBTAINABLE FROM THIS COMPACT LOW PRICED SYSTEM

PAIR OF LOUDSPEAKER UNITS

incorporating high flux  $8 \times 5$  ins. speaker. Size approx.  $13 \times 7\frac{1}{2} \times 8\frac{3}{4}$  ins.

Price complete £42

#### LOW DEPOSIT **CREDIT TERMS AVAILABLE ON PURCHASES OVER**

£8 (KITS OF PARTS EXCEPTED)

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	6AL5	.11	6L18	. 45	12J7GT	.50	35W4	.23	E92CC	.40	EF36	.33	HL42D		PEN360		t
	6AN8	.49		38	12K5 12K7G		35Z4GT	.24	E180F	.95	EF37A	.45	HN309		PEN45	.35	l
	6AQ5	. 22	6LD20	.48	12Q7G7		35Z5GT	.30	E182CC		EF39	.40	HVR2	.53	PEN45I		ì
	6AT6	.20	6N7GT	.40			50L6GT	.45	E1148	.53	EF40	.50	HVR2A	.53	PEN451		ļ
×	6AU6	. 21	6Q7	. 43	12SA76	.40	85A2	.48	EA50	.18	EF41	.58	KT2	. 25	PEN46	.75	ł
	6AV6	.30		.30	12SC7	.35	90C1	.59	EA76	.88	EF42	.33	KT8	1.75			1
	6B8G	.13		. 55		. 23	807	.59	EABC8		EF54	.98	KT44	1.00	PEN453		1
	6BA6	.21	6R7G			.15	5763	.50		.38	EF80	. 23	KT66	.82		. 98	ì
	6BC8	. 50		.35		. 23	AC2/PE		EAF42	.49	EF83	.48	KT81	2.00	PENDL	-	ŀ
	6BE6	.22			12SK7	. 24	HOLIT I	. 98	EB34	.20	EF85	.28	KT88	1.70	4020	. 88	ı
	6BH6 .	.43					AC2/PE		EB91	.11	EF86	.31	KTW61		PFL200		ļ
	6BJ6	. 39		.53	12061	. 50	DD	. 98	EBC41	. 48	EF89	.24	KTW62	. 63	PL36	. 47	1
	6BQ7A	,38	68H7	. 35	1487	1.15	AC6PEN		EBC81	.33	EF91	.17	KTW63	.50	PL81	. 44	1
	6BR7	.79	68J7			. 24	AC/PEN		EBC91	.30	EF92	. 35	MHLDe	.75	PL81A	. 50	١
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	6BW6	.72		.17		.66	AC/TH1			.38	EF98	. 65	N308	.98	PL83	. 33	i
	6BW7	.57		.31	20L1	.98		.98	EBF89	.30		.27			<u> </u>		
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						1.00		.46		.35	EL34	.46	damage	in tra	insit for o	nly 0	۱.(
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	6CH6	.38		.35		.22		.91		1.58	EL37	.87	We do 1	not ha	andle seco	nds:	n
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	€CU5	.50	7H7	. 60	12010	.00	10220					_				_	-

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	TT 10	ro .	N339 .44	PL84 .31	UCL83	.49	AC107	.15	BCZ11	38	OA70	.15
		.53		PL302 .60		.50	AC113	.25	BD119 .	45	OA73	.15
		.50	P61 .49	PL504/500		. 60	AC127	.17	BF158	29	OA79	.09
		.38	PABC80 .34	*84		.35	AC128	.20		25	OA81	.09
		.23	PC86 .49				AC156	.20		20	OA85	.08
	EL85	.40	PC88 .49	PL505 1.30		.34						.13
		.40	PC95 .53	PL508 .90		. 63	AC157	. 25	BF173	.38	OA90	.10
W.1		. 23	PC97 .38	PL509 1.30		.29	AC166	. 25			OA91	. 09
90	EL95	.34	PC900 .35	PL802 .75		.59	AC168	.38		.40	OA95	. 09
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.17		.38	PCC85 .29	PX4 1.16	UM80	.33	AC177	.28	BF194	.15	OA202	.10
		.40	PCC88 -44	PX25 1.16	UYIN	.50	ACY17	.25	BFY50	. 23	OC22	. 38
.20		.32	PCC89 .46	PY33/2 .50	UY21	.55	ACY18	.20	BFY51	.19	OC23	.38
.23				PY80 .33		.38	ACY19	.19	BFY52	.20	OC24	.38
. 29		. 36		PY81 .25	UY85	.26	ACY20	.18	BY100	.18	OC25	.38
.27		.34		PY82 .25	U10	.45	ACY21	.19		.18	OC26	.24
.40		. 35	PCF80 .29			.38	ACY22	.15		.18	OC28	.60
.35		. 55	PCF82 .31		U18/20	.75	ACY28	.18		.15	OC29	. 63
9 .48	EY84	.50	PCF84 .40	PY88 .33				.36		.18	OC35	.32
4 .57	EY87/6	.31	PCF86 .45	PY500 .95		.73	AD140			.10	OC36	.43
71.70		.43	PCF200 .67	PY800 .35		.39	AD149	.50	BYY23			
.28	EY91	.53	PCF801 .32	PY801 .34	U25	.65	AD161	.45		.00	OC43	.63 1.18
.28		.40	PCF802 .40	PZ30 .48		.59	AD162	. 45	BYZ10	.25		.10
.65	EZ41	.43	PCF805 .63	QQV03/10		. 78	ADT140		BYZ11	.25	OC44	
42.10		. 22	PCF806 .59	1.20		.60	AF102	. 90	BYZ12	.25	OC45	.11
1 .63	EZ81	.23	PCF808 .70	QV04/7 .63		.70	AF106	.50	BYZ13	. 25	OC46	.15
2 .62	EZ90	.22	PCH200 .62	RG1-240A	U281	.40	AF114	. 25	CG12E	.20	OC70	.18
1 .28	FW4/500		PCL82 .33	1.98	U282	.40	AF115	.15		. 23	OC71.	.11
3 .40	FW4/800	75	PCL83 .60	R10 .75	U301	.40	AF117	.19	GD9	.20	OC72	.11
4 .36		.35	PCL84 .35	R11 .98	U403	.33	AF121	. 30	GET113		OC74	. 23
				R16 1.75	U404	.38	AF124	. 25	GET116	.40	OC75	.11
.32	GZ-02	.42	PCL805/85	R17 .88	U801	. 95	AF126	.18	GET118	.20	OC76	.15
.81			.42	R19 .31	U4020	.38	AF139	. 65	GET119	.20	OC77	. 27
.52	GZ34	. 53	PCL86 .40	SP42 .75	VP23	.40	AF178	.68	GET573	.38	OC78	.15
.55		.75	PCL88 .67	SP61 .33	VP41	.38	AF180	.48	GET587	.43	OC78D	.15
.55	HABC80		PD500 1.44		VR105	.33	AF186	. 55	GET873	.15	OC81	.11
.37	HL23DD		PEN4DD		VT61A	.35	AF239	.38	GET887	.23	OC81D	.11
. 63	HL41DD		1.38	TH233 .98	VUIII	.44	BA102	.45	GET897	.23	OC82	.11
.33	HL42DD		PEN36C .75	TP2620 .98	VU120	.60	BA115	.14	GET898	.23	OC82D	.11
4 .45	HN309 1		PEN45 .35	UABC80 .30	V U 120			.25	M1	.15	OC83	.20
.40	HVR2	. 53	PEN45DD	UAF42 .50			BA116 BA129	.13	M3	.15	OC84	.24
.50			.75	UBC41 .45	VU133	.35		.10	MATI00	. 13	OC123	.23
. 58	KT2	. 25	PEN46 .20	UBC81 .40	W76	.34	BA130	.10	MATIOO	.39	OC139	.23
.33		75		UBF80 .29	W107	.50	BA153	.15	MAT101	.09	OC140	. 95
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.28	KT88 1	.70	4020 .88	UCC84 .34	& Diodes		BC115	.15	38 A M 103	. 39	0C200 0C201	.38
.31	KTW61		PFL200 .54	UCC85 .35			BC116	. 25	MAT121	40		. 43
.24		.63	PL36 .47	UCF80 .35	28323	.50	BC118	.23	0.17	.43	00202	.30
.17		.50	PL81 .44	UCH21 .60	AA119	.15	BCY10	. 45	OA5	.28	OC203	
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3Q4 0.4 3S4 0.3		6N7GT 0.40	12B4AA0.55	30L1 0.40	866A 0.75	E88CC 0.65	ECF82 0.35		0.43	PCC88 0.55	PY82 0.30	UCH21 0.60
3V4 0.4		6P1 0.60	12BA6 0.35 12BA7 0.35	30L15 0.85 30L17 0.80	5642 0.65	E180F 0.95	ECF86 0.65		0.40	PCC89 0.50	PY83 0.38	UCH42 0.70
5R4GY 0.6		6P25 1.25	12BE6 0.35	30P12 0.80	6080 1.50 6146 1.50	E810F 2.90 EABC800.35	ECF8041.50 ECH42 0.70		0.35	PCC189 0.55	PY88 0.40	UCH43 0.75
5U4G 0.3		6P28 0.65	12BH7 0.40	30P19 0.80	6146A 2.00	EAF42 0.55	ECH81 0.30		0.35 1.00	PCC805 0.85 PCC806 0.80	PY500 1.00 PY800 0.50	UCH81 0.35 UCL81 0.60
5V4G 0.4		6Q7 0.40	12BY7 0.55	30PL1 0.70	6146B 2.50	EBC33 0.50	ECH83 0.40	EL821		PCF80 0.80	PY800 0.50 PY801 0.50	UCL81 0.60 UCL82 0.35
5Y3GT 0.8		68A7 0.40	12K5 0.55	30PL13 0.93	6360 1.25	EBC41 0.55	ECH84 0.45		0.90	PCF82 0.35	PZ30 0.35	UCL83 0.60
5Z3 0.5	0 6CG7 0.50	68G7 0.35	12K7GT0.35	30PL14 0.90	6870 2.50	EBC81 0.30	ECL80 0.40	ELL80		PCF84 0.50	QQV2-6 2.15	UF41 0.60
5Z4G 0.4		68K7 0.35	12Q7G 0.80	35A3 0.55	6939 2.15	EBF80 0.40	ECL81 0.45		0.90	PCF86 0.60	QQV03-10	UF42 0.60
6/30L2 0.7		68L7GT0.35	128R7 0.35	35A5 0.75	7199 0.75	EBF83 0.40	ECL82 0.35		0.75	PCF87 0.85	1.25	UF43 0.60
6AB4 0.3		6SN7GT0.35	1487 0.80	35B5 0.65	7360 1.80	EBF89 0.30	ECL83 0.65	EM80 .	0.40	PCF801 0.50	QQV03-20A	UF80 0.35
6AF4A 0.5		68Q7 0.40	20D1 0.45	35C5 0.40	7586 1.25	EC53 0.50	ECL84 0.55		0.60	PCF802 0.50	5.25	UF85 0.40
6AG7 0.4		6SR7 0.40 6T8 0.35	20L1 1.10	35D5 0.70	7895 1.25	EC86 0.60	ECL85 0.55		0.35	PCF805 0.75	QQV06-4A	UF89 0.35
6AH6 0.5 6AJ8 0.3		6U4GT 0.60	20P1 0.50 20P4 1.10	35L6GT 0.50 35W4 0.30	9002 0.35	EC88 0.60	ECL86 0.40		1.00	PCF806 0.70	5.50	UL41 0.65
6AK5 0.8		6U8A 0.40	20P4 1.10 20P5 1.20	35W4 0.30 35Z3 0.60	9003 0.50 AZ1 0.48	EC90 0.33 EC91 0.50	ECLL800		0.55	PCF808 0.75	TT21 2.65	UL84 0.30
6AK6 0.5		6V6GT 0.38	25C5 0.50	35Z4G 0.80	AZ1 0.48 AZ31 0.55	EC91 0.50 EC92 0.35	1.50 EF37A 0.60		0.35	PCH2000.70	TT22 2.80	UM84 0.20
6AL3 0.4		6X4 0.80	25L6GT 0.45	35Z5GT 0.40	CBL1 0.80	EC93 0.50	EF37A 0.60		0.40	PCL81 0.50 PCL82 0.35	U18/20 0.75 U25 0.75	UY1N 0.50 UY11 0.65
6AL5 0.2		6X5GT 0.35	25Z4G 0.30	50A5 0.70	CBL31 0.90	EC8010 2.25	EF40 0.50		0.40	PCL83 0.65	U25 0.75 U26 0.75	UY11 0.65 UY41 0.45
6AM5 0.3		6X8 0.55	25Z6GT 0.65	50B5 0.45	CY31 0.35	ECC40 0.60	EF41 0.65		0.55	PCL84 0.45	U31 0.45	UY82 0.50
6AM6 0.3		6Y6G 0.65	30A5 0.45	50C5 0.40	DAF96 0.42	ECC81 0.33	EF42 0.70		0.40	PCL85 0.40	U37 1.50	UY85 0.30
6AQ5 0.3	5 6EW6 0.65	7Y4 0.60	30AE3 0.40	50CD6G1.65	DF96 0.42	ECC82 0.30	EF80 0.25		0.43	PCL86 0.45	U52 0.38	W729 0.60
6AQ6 0.5	5 6F1 0.70	9BW6 0.50	30C1 0.80	50L6GT 0.50	DK40 0.55	ECC83 0.30	EF83 0.55		0.43	PCL88 0.90	U76 0.30	Z759 1.85

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2N706	0.15	AC126	0.25	BCZ11	0.40
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2N1305	0.25	ACY17	0.30	BF197 BFW87	0.20
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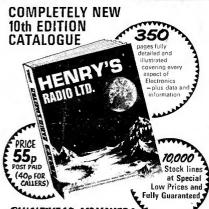
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