## PRACTICAL <br> WRE <br> ? <br> NOVEMBER 1968

REE INSIDE

- AUDID REFERENGE CHABT TABLES, DATA,


## GOLOUR GODES, ETG,ETG

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(9)

## Stereo cartridge connections

There are many different pin and tag formations on gramophone pickup cartridges. Fig. 1 brings together most of the popular configurations These are stereo cartridges, with left channel live connection marked 'L', left earth, 'LE', right channel live, ' $R$ ' right channel earth 'RE' and common connection where this applies marked ' C '. For monaural connection, it is usual to parallel channels, connecting the live pins together and the earth pins together unless these are already commoned.
The diagrams show the rear view of cartridges, with stylus in the playing position Dotted lines indicate fitted links. In most cases it is necessary to link at the aniplifier, rather than the cartnidge end
Types of cartidges using the connections shown are principally as follows.
(a) (b) (c) and (d) Acos (Cosmocord)
(e) ADC (KEF Electronics)
(f) BSR, (also 4 -pin type as in (k).)
(g) Connoisseur
(h) Decca Mark I and II
(1) Decca Deram
(1) Decca Deram (crystal and ceramic evpes for universal application have turnover styli, shown universal application have turnove
() Elac 'plug-in' types, crysial and ceramic
(1) Elac plug-In iypes
(m) Empire (De Villiers)
(n) Ortofon (Metro-Sound) link for mono show
) and (p) Philips. with shorting link for mono shown dotted. (p)
(q) Pickering (Goldring)
(r) Ronette (Some types are three-pin, with left pin for left channel, right pin for right channel and centre pin for common connection)
( $s$ ) and ( t ) Shure. (Pin spacings may differ, but phasing is as shown.)
(u) Shure (see above)
(v) Tannoy.
(w) Garrard (five different plug-in head shells are employed and wiring to cartridges is colour-coded. Earlier versions used left grey. left-earth, blue, right, brown, right-earth or common. black Later versions conform to international practice, left, white, left-earth. blue right, red, right-earth, green.
(x) Goldring

## Plugs and sockets

DIN standards resolve to 5 types of plug and so pin formation, as in Fig. 3
Standard connections t, $5-$ pin DIN socket for at purposes are as shown in Fig. 4. This is for st recording and replay. Vaiations permit parallel p crossing of connections 'or dubbing and additio alternative sources.
Phono plugs may hav long or short pin leng Cinch (Carr Fasteners Lid) plugs may have shorter length. Some Japanese iynes are also shorter than u British version. Care must lie taken to ensure connec British version. Care must he taken to ensure connec and to avoid short-circuit. caused by using the wr
length plug. ength plug.
Jack plugs are in 4 main variettes. classed accor to diameter. 25 mm is miniature version, gene employed for headphone; etc 3.5 mm is freque
used input connection, i.specially on Japanese used input connection, ispecially on Japanese
recorders. Also employed for foot controls, head recorders. Also employed for foot controls, head etc.
5 mm is special diamiter, limited in use to $F$ and Luxor.
nconvan

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Jack plugs are in 4 mal varıetıes, classed accordıng diameter. 25 mm is miniature version, generally iployed for headphones, etc 35 mm is frequently ed input connection, ispecially on Japanese tape corders. Also employec for foot controls, headsets,

5 mm is special diamiter, limited in use to Ficord d Luxor.

(3)

PN 2 is alwars Eantm or chassis cenmectiom


Standard 1-in jack (G.PO type) has tip same diameter as barrel, Stereo jack is similar to standard with additional ring. Tip may be of smaller diamete than sleeve. Tip is normally Channel 1 , left-hand, and ring Channel 2, right-hand


## Audio spectrum

Frequency ranges of some commonly heard instruments and voices. Considerable overtone modification may be found, affecting the timbre, and giving instruments their characteristic sound

## Metric equivalents

| Gramophone disc sizes | $\begin{aligned} & 12 \mathrm{~mm}-30 \mathrm{~cm} . \\ & 10 \mathrm{mn}-25 \mathrm{~cm} . \\ & 7 \mathrm{in}-175 \mathrm{~cm} \end{aligned}$ |
| :---: | :---: |
| Gramophone stylus sizes | $1 \mathrm{mll}-25 \mu$ |
|  | $07 \mathrm{~mol}-18 \mu$ |
|  | 05 mil - $13 \mu$ |
|  | $0.2 \mathrm{mll}-5 \mu$ |
| Tape speeds. | $\begin{aligned} & 15 \mathrm{in} \mathrm{sec}-381 \mathrm{~cm} / \mathrm{s} . \\ & 7 \frac{1}{2} \mathrm{in} \mathrm{sec}-191 \mathrm{~cm} / \mathrm{s} \end{aligned}$ |
|  | $33 \mathrm{ln} / \mathrm{sec}-95 \mathrm{~cm} / \mathrm{s}$. |
|  | 17. $\mathrm{ln} / \mathrm{sec}$. $-4.8 \mathrm{~cm} / \mathrm{s}$ |
|  | $13 \mathrm{in} / \mathrm{sec} .-24 \mathrm{~cm} / \mathrm{s}$. |
| Tape spool sizes | $7 \mathrm{~m}=17.5 \mathrm{~cm}$. |
|  | 53 ln . -14.6 cm . |
|  | $5 \mathrm{~mm} .-12.7 \mathrm{~cm}$. |
|  | $4 \mathrm{~m} .-10 \cdot 2 \mathrm{~cm}$. |
|  | $3 \mathrm{~m} .-76 \mathrm{~cm}$. |

Atmospheric pressure:
normal 14 7lbs, sq.in. - 1.034 Kgrnsisq.cm.
Sound velocity at 20 deg . C.
$1130 \mathrm{ft} / \mathrm{sec}$. -344 metres $/ \mathrm{sec}$.
Sound velocity at 0 deg. C.
$1087 \mathrm{ft} / \mathrm{sec} .-331$ metres $/ \mathrm{sec}$.

for left channel, right pin for right channel and centre pin for common connection.
(s) and (t) Shure. (Pin spacings may differ, but phasing is as shown.)
(u) Shure (see above)
(v) Tannoy.
(w) Garrard (five different plug-in head shells ar employed and wiring to cartridges is colour-coded Earlier versions used left grey, left-earth, blue right, brown, right-earth or common, black. Later versions conform to international practice, left white; left-earth, blue right, red right-earth, green (x) Goldring
(y) Sonotone.
(z) Orbit (Neat)

Sony cartridges employ basic pin connections, in line, left, left-earth, right-earth and right

B \& O cartridges are as in (aa)
Audio-Technica use two upper pins left and right with conforming lower pins linked, and commoned


(3)

(1)

(d)



EARTM ME
(8)
 inverts

(9)

## Pick-up armand shell connections and colour codes

International agreement for pick-up cartridge and arm colour-coding is now as follows
Left channel live - white
Left channel earth - blue
Right channel live - red
Right channel earth - green
Common or screen - black
There are numerous variations, and earlier models may have grey brown and black wirings for left righ and common connections. See Fig. 2 ( f ) and ( g ).

Connections to shells and arms of some popular ariations are as shown below
(a) B.S.R tag-board wiring. With three-pin systems, blue is not used
(b) and (c) Ortofon pick-up arm connections (Colour coding may be non-standard, with red left and white right channel, and blacks for both earths. A separate black may also be used for connection to the arm.)
(d) SME shell connections. Colour coding is 1 , red 2. green; 3, yellow and 4, black. Left channe connects to 1 and right channel to 2 , with earth connection to 4.
(e) SME connections to tag strip.


## Operational symbols

Despite many variations by different manufa some general agreemert on international synit audio purposes has evol.ed. Fig 5 lists those in use. Note that the symbul for remote control ni denote 'Tape recorder sncket' on some amplifie that for Input/Output may be a simple output so some tape recorders and associated equipment

Gramophone record sta

> Disc Sızes:
> Playing speeds

Grooves per ini $h$
Recorded diameters:
$11 \frac{2}{8}=\frac{1}{32}$ in. 112
$9 \frac{1}{8}=\frac{1}{12} \ln .(10$

$75 \cdot 11=5 \%(45)$
$45.11=5 \%(78)$ $33 \cdot 33=5 \%(33)$
(78 r.p.m.) coarse
( 33 and $45 \mathrm{r} . \mathrm{p} . \mathrm{m}$
Outer-12-in. $11 \frac{1}{2}$
$\begin{aligned} 10-\mathrm{in} . & 91 \\ 7-\mathrm{in} . & 612\end{aligned}$
Inner-12-in. $43^{32}$
$10-\mathrm{in} .4 \frac{1}{4}$
7 -in. $3_{4}^{3}$
Centre hole dismeter: $0.285-0.2885 \mathrm{in}$
(for 45 r.p.m. dis
1-502-1-506 in.)

For compatibility between tape recorders and to enable interchange of tapes, exact dimensions of recorded tracks are needed There are several standards the requirements of which overlap to some extent, with resultant changes in tolerance and in quality

Tape width for domestic purposes is taken to be one quarter-inch, but should actually be 0246 in . with a maximum permissible tolerance of $=2$ thou ( 0002 in .) Earlier standards specified $\frac{1}{4}$-inch $0-6$ thou

Fig. 6 shows some of the principal dimensions, with safety lanes between tracks and at edges of tape also given. Track width is determined by the gap dimen sions: some iwo-track tape recorders use 0.118 in track width and have their heads adjusted to scan the edge of the tape. Safety margins between tracks are reduced to as little as 001 in Two-track stereo heads with rack widths of 008 in are used giving a wide watety margin and less cross-track interference but saleil dropouts. Four track recording is more general for prerecorded stereo tapes
(a) half-track recording
(b) half-track recording with reduced track width showing direction of recorded information.
(c) quarter-lrack recording, showing safety lanes
(d) quarter-track recording, showing direction of recorded information For stereo recording, track 1 is left channel, track 3 right channel, in one direction, inversion of tape then makes track 4 left channel and track 2 right channel.
(e) metric dimensions, half-track to the left, showing method of overlap to procure full advantage of rack and reduce vartitions due to tape and trans port discrepancies. Quarter-track recording may also be carried to the edge of tape, as shown on the right of the diagram.
manufturers me general agreemem on international symbols fo purposes has evolved. Fig 5 lisis those in popula e. Note that the symbol for remote control may also Tape recorder socket on some amplifiers, ano al ome tape recorders and associated equipment.

## Gramophone record standards

Disc Sizes

Playing speeds

Grooves per inch:
Recorded diameters
$11 \frac{2}{8}=\frac{1}{51} \mathrm{in}$ ( $\left.12-\mathrm{in}.\right)$

| $9 \frac{7}{8}=\frac{1}{J_{2}} \ln .(10-\mathrm{in})$. |
| :--- |
| $\frac{1}{2} \ln .(7-\mathrm{in})$ |

$68 \quad \frac{1}{32}$ in. $(7-\mathrm{in}$.
$77.92-5 \%(78)$ r.p.m.
$45 \cdot 11=5 \%(45)$ r.p.m.
$33.33=5 \%$ (33) r.p.m.
( 33 and. 45 r.p.m.) fine, 200-300
Outer-12-in. 11 $\frac{1}{2}-1 n$.

$$
\begin{array}{cl}
10 \text {-in. } & 93 \text {-in. } \\
7 \text {-in. } & 6 \frac{1}{32}-\mathrm{in} .
\end{array}
$$

Inner-12-in. 43
10 -in. $4 \frac{1}{3}$
7 -in. $3 \frac{3}{4}$
Centre-hole diameter: $0.285-0.2885$ in
(for 45 r.p.m. discs, optionally
1.502-1.506 in.)

Eccentricity: centre of centre-hole not more than 0.005 in. from centre of groove spiral
Siylus radius: Mono fine-groove $1 \mathrm{mil}(25 \mu)$, stereo $0.5 \mathrm{mil}(13 \mu)$
Bi-radial (elliptical) styli. Major radius 0.70 .8 mil minor radius 020.3 mil
Disc Replay Standards: for pickups having an electrica output proportional to stylus velocity (magnetic) mean slope of curve is 4 dB per octave. (Response approximately $13 \frac{1}{2} \mathrm{~dB}$ up and down at $100 \mathrm{c} / \mathrm{s}$ and $10 \mathrm{kc} / \mathrm{s}$.) Crystal and ceramic pickups usually have response falling at 6 dB per octave when fed into high resistive load.

## Normal recorded velocity: $5 \mathrm{~cm} / \mathrm{sec}$

Stereo colour-code: Left-white; left-earth-blue:
Right-red; right-earth-green
Vertical tracking angle: 15 degrees
$1 \mathrm{Bm} / \mathrm{sec}-4.8 \mathrm{~cm} / \mathrm{s}$

Aimospheric pressure
normal 14 7lbs/sq.ın. $-1.034 \mathrm{Kgms} / \mathrm{sq} \mathrm{cm}$ Sound velocity at 20 deg . C
$1130 \mathrm{ft} / \mathrm{sec} .-344$ metresisec at 0 deg . C .
$1087 \mathrm{ft} / \mathrm{sec}$. -331 metres/sec
(rising approximately 2 fi/sec. per deg. C.)
Acceleration: Standard g. 32 ft . per sec. per sec. 980 $\mathrm{cm} / \mathrm{s}$ /s


## Conversions

| 1 inch | - 2.54 cm . |
| :---: | :---: |
| 1 thou | - $25.4 \mu$ |
| 1 foot | - 30.48 cm . |
| 1 sq.in. | - 6.452 sq.cms. |
| 1 cuin. | - 16.39 cu.cms. |
| 1 gram | - $980 \cdot 62$ dynes |
| 1 oz . | - 28.35 gms . |
| 1 lb . | - 0.4536 Kgms . |
| $1 \mathrm{lb} / \mathrm{sq} . \mathrm{ın}$. | - $70.307 \mathrm{gms} / \mathrm{sq} . \mathrm{cm}$. |
| 1 micron ( $\mu$ ) | - 0.0394 thou (.00004 in.) |
| 1 cm . | - 0.3937 in. |
| 1 metre | - 39.3708 in . |
| $1 \mathrm{sq.cm}$. | - 0.155 sq in |
| 1 cucm | - 0061 cu in |
| 1 dyne | - 0.00102 gramme weight. |
| 1 watt | - 0.00134 horse power |
|  | - $44 \mathrm{ft.lbs} / \mathrm{min}$ |

1 thou $\quad 25.4 \mathrm{~cm}$
1 foot -30.48 cm
sa.in - $6.452 \mathrm{sq} . \mathrm{cms}$
1 gram - 980.62 dynes
oz.
lb/sq.in. $-70.307 \mathrm{gms} / \mathrm{sq} . \mathrm{cm}$.
cm . $\mu$ ) 0.3937 inou (. 0004 in .)
$39 \cdot 3708$ in
1 sq.cm. $\quad-0.155 \mathrm{sq}$ in
1 dyne - 0.00102 gramme weigh
$-44 \mathrm{ft} . \mathrm{lbs} / \mathrm{min}$

- $0.1 \mathrm{Kgm} / \mathrm{m} / \mathrm{sec}$


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SAVBIT

## SIZE 5

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## SIZE 12

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TONE CONTROLS ARE COMMON TO ALL INPUTS
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Sensitivity: M. K. at $1 \mathrm{Mc} / \mathrm{s} 10$ microvolts plus or minus 40 . - Kigh $Q$ interizal ferrite or minus 4 dB

- Class 'B' modul

Class 'B' output stage ensures iong battery listor controlled heat stabilisation to the output level. Total current drain of the receiver under no signal conditions is $10-12 \mathrm{~mA}$. At reasonable listening level $20-30 \mathrm{~mA}$. - Extension socket for car aeria! iaput, tape recorder output (
of volume control) and External Bpeaker.

- All components (except apeaker) mount oa the printed circuit boari1. Easyto follow inatructions. Size of cabinet 12 m . long, 8 in . high and 3in. deep
Finger-tip contrsls. - Flager-tip contrss.

Special Offer-Power Supply Kit to purchasers of Dorset Portable Radio parts incorporating mains translormer, rectifier and smoothing condenser, AC mains $200 / 250 \mathrm{~V}$ output $9 \mathrm{~V} 100 \mathrm{~mA}, \mathrm{~B} / 6$ extra

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lag mains tranaformer. rectlier sind smoothing 20)/250 volts. O Mitput 20)/25 9/6 exirs 10 mp Plus $7 / 6 \mathrm{P}$. \& $\mathbf{P}$. Parts circuit diagran 2/6 FREE with parts.

- De luxe wooden cabinet ata $12 \frac{1}{2} \times 8 \frac{1}{\prime \prime}^{\prime \prime} \times 3 \frac{1}{n}^{\prime 2}$.
$\star$ Horizontal easy to read tuning scale printed grey with black

* High 'Q' ferrite rod aerial
* I.F, neutralization on each separate stage.
t D.C. coupled push pull output atage with separate A.C. negative feedback.
* Room flling output 300 mW
* Ready etched and drilled printed circuit board back printed for foolproof construction.
* Fully comprehenrive instructions and point-to-point wiring diagrams
* Car aerial socket.
* Fulty tunable over medlum and long wave, 168-536 metres and 1250-2000 metres.
* All components ferrite rod and tuning assembly mount on printed bonrd
* $5^{\prime \prime}$ P.M. вpeaker
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＂PACKAGE 3＂＇ 30 WATT SYSTEM ＊Goldring Transcription Turntable o ＊Super 30 Amplifler in Teak veneer housing ＊Pair of Stanton Loudspeaker Units Special inclusive price．Fully wired 82 Gns． units ready to＂plug－in＂＇
TERMSAVALLABLE ALLACKAGE OFFERS AUDIOTRINE HIGHFIDELITY LOUDSPEAKERS

Heavy cast construction．Latest high efficiency round giving low fundamental reso nance．＂D＂indicates Tweeter Cone providing extended frequency range． Impedance 3 or 15 ohms．Please state choice．Response $40-18,000$ c．p．s．Ex ceptional performance at low cost
 Prices include carriage

## HF501




HIGH FIDELITY LOUDSPEAKER UNITS Cabinets of latest styling Satin Teak or Walnut acoustically lined（and ported where appro priate）．Credit terms avallable on all units

 STANTON IIIS Size $18 \times 11 \times 10 \mathrm{inc}$ ．Rating STANTON IIIS Size $18 \times 11 \times 10 \mathrm{inc}$ ．Cating 10 watts．Incorporating Audiotrine HF 815 speaker with roll rubber surround and15．000inemagnet．High fux tweeter Response $30-20.000 \mathrm{c}$ ．．p．s．Impdnce 3 or 1516 Gns ．
ohms．Glvessmooth realisticsound output． GLOUCESTER 12.000 line speaker．Cross－over unit and ${ }^{2}$ Weeter Rating 10 watts．Smoth response $12 \frac{1}{2}$ Gns．
E2 EQUIPMENT CABINET
Size $17^{\prime \prime}$ wide $14 t^{*}$ deep $11 t^{\prime}$ high and other amplifiers．＇Hinged＇ perspex cover．Satin Teak veneer finish． 8 Gins． MOTOR BOARDS Garrard Turntables and $12 / 9$ R．S．C．TA6 6 Watt HIGH FIDELITY SOLID STATE AMPLIFIER

$$
Q^{6}
$$ Freauv．A Frequency $20,000 \mathrm{c.p.s}$ Response Distortion $0.3 \%$ at 1,000 c．p．s lift＇and＇cut＇controls． 3 input sockets for Mike． Gram，Radio or Tape．Input selector switch．Output for $3-15$ ohm speakers．Max sensitivity 5mV．Fully brushed sllver finish facia plate $10 t x 3+1 n$ ．and matching knobs．Complete kit of parts with full wiring diagrams and instructions．Carr．7／6

Matched for optimum periormance．Sub－ individually．


Illustrated with TFM1 Mk II Tuner fitted EXTREMELY ATTRACTIVE AND VERSATILEP LINTHS flnished in Satin Teak veneer．TintedP erspex RECORD PLAYING UNITS Money saving units．Ready to RP2 plug into Amplifiro sisting of Garrard SP25）Mk IT Goldring CS90 high compl ance ceramic Stereo／Mono cartridge with diamond sty－ Lus．Mounted on Plinth Pers
pex Cover 3 gns． 19 Gns pex Cover 3 gns． 19 Gns． hT As above but with Goldring Lenco GL88 Trans ription unit and CS90 Cart
ridge．Perspex Cover 3 gns ana fiti 24 Gns．


3000．AT6，AT60．SP25 or Gold ring GL68．Available with clear Perspex co－ 6 Gns． Ver as ill．Inc．Carr． at 3 Gins．Limited number of covers slightly damaged but
repaired by makers． $39 / 9$

## INTEREST CHARGES

 REFUNDED on Credit Sales settled in 3 months ．$\begin{array}{lr}\text { hinged cover } \\ \text { with } & \text { satin } \\ \text { chrome handie．}\end{array}$

$\sigma_{\infty}^{\infty}$JE8 Size $16 \times 11 \times$ in．Gives
with any 8 in．Hi－Fi speaker．
＋Garrard SP25 Mk II Turntable on Plinth ＊Goldring CS90 Ceramic P．U．Cartridge $\star$ Super 30 Amplifer in Teak veneer housing ＊Pair of Stanton Loudspeaker Unit Special inclusive price．Fully $\quad 72$ Gns．
wired units ready to＂plug－in＂．
＂PACKAGE 1＂ 13 WATT SYSTEM ＊Garrard SP25 Mk II 4 speed Player Unit $\star$ Goldring CS90 Ceramic P．U．Cartridge ＊TA12 Amplifier in Teak veneer housing ＊Pair of Dorset Loudspeaker Units Special inclusive price．Perspex cover 59／9 extra． Or Dep．£8．18．6 and 9 monthly
payments £5．3．0．（Total £55．5．6）． 48 Gns．

## REC．CTAR IS WATT STERED AMPIEJER

FULLY TRANSISTORISED，SOLID STATE CONSTRUCTION HIGH FIDELITY OUTPUT OF 6.5 with any crystal or ceramic Gram corder．＇Mike，etc．$\star 3$ separate switched input sockets on each chan－ nel $t$ Separate Bass and Treble con Speal $t$ Slide Switch for mono use $\star$ 200－250v A．C．mains $\star$ Frequency Response A．C．mains $30-20$ Frequency c．p．s．Hum and Noise－ $70 \mathrm{~dB} \star$ Sensitivities（1） 300 mV （2） 50 mv （3） 100 mV （4） $2 \mathrm{mV} *$ Handsome brushed silver finish Facia and Knobs Complete kit of parts with full wiring diagrams and in－ 17 Carr Deposit 24.16 .0 and 9 mthly pymts． 28 g－（Total 17 GNS ）． $\mathrm{C}_{\text {GNS }}$ AUDIOTRINE HI－FI SPEAKER SYSTEMS
Consisting ofmatched $12 \mathrm{in} .12,000$ line 10 watt 15 ohm high quality spesker，cross－over unit and tweeter號 sure surprisingly realistic reproduction． 5 GnS

HI－FI SPEAKER ENCLOSURES Teak veneer finish Modern design．Accoustically lined and ported
Modern design，Accoustically lined and ported．
SE8 For optimum performance
Hi－Fi speaker．Size $22 \times 15 \times 9$ in． ． 22 For outstanding results SEl2Forellent with 10 in Hi－Fi＇spkr 5 Gns． $12 i n$ Hi－Fi＇spkr and Twee－ 6 Gns．
Size $24 \times 15 \times 10 i n$ ．

## R．S．C．TFM1 SOLID STATE VHF／FM RADIO TUNER



High－sensitivity t zoo－zs0v．A．C．Malns opera－ ton．太 Sharp A．M．Rejection．$\star$ Drift－free recep－ tion．A Output ample for any amplifler（approx． $500 \mathrm{~m} . \mathrm{v}$. ．太 Simple alignment instructions．＊Out－ put available for feeding tuning meter．© Output using silicone Planar Transistors．Tuner head for standard 80 ohm co－axial input．Visually matching our Super 15 and for standard 80 ohm co－axial input．Visualy matching our super 15 and The pre－wired tuning head facilitates speed and simplicity of construction Printed circuitry．Only first grade transistors and components used． quality product at half the cost of comparable units．Stereo version，all quality product at half the cost of comparable unit
parts 18 gns．Carr．10／．Assembled $22 t \mathrm{gns}$ ．inc．carr．

##  FULLY THANSISTUKISED 200／250v．A．C．Mains OUTPUT 10 WATTS R．M．S．cont．into 15 ohms． SPECIFICATIONS COMPARABLE A DUAL CHANNEL VERSION OF THE SUPER 15， Gmploying Twin Printed Circuits．Close tolerance -52 dB at $1,000 \mathrm{c}$ p．s． CONTROL： 5 position Input Selector，Bass Control．Treble Control．Volume Control．Balance Control．Stereo／Mono Switch．Tape Monitor Switch．Mains．Switch．INPUT SOCKETS Matched Pairs）．（1）Magnetic P．U．（2）Ceramic or SOCKETS（Matched Pairs）．（1）Magnetic P．U．（2）Ceramic or Crystal P．U．（3）Radio／Aux．（4）Tape Head／Microphone． Operation of the Input Selector Switch assures appropriate equalisation．Rigid $18 \mathrm{~s} . \mathrm{w} . \mathrm{G}$ ．Chassis．Size approx．12tn．wide， ain．high and 81 n ．deep．Neon Panel indicator．Attractive Fe－ 3in．high and 81n．deep．Neon Panel indicator．Attractive Fa－ 2.5 mV ．Radio／Aux or Ceramic P．U． 110 mV ．Tape Head <br>  LATEST TRANSTSTORA．S．cont，into $3-4$ ohms． NKT713，NKT217，NKT217．NKTT275，NKT275，NKT274， NKTT13，NKT217，NKT217，NKT717，NKT403， EQUALISATION to Standard R．I．A．A．and C．C．I．R． FULL TAPE MONITORING FACILITIES． SENSITIVITIES：Masnetic P．U． 4 mV ．Crystal or FREQUENCY RESPONSE： $\pm 2 \mathrm{~dB} 20-20,000$ c．p．s． TREBLECONTROL：+15 dB to -14 dBa at $10 \mathrm{Kc} / \mathrm{s}$ ．NEG FEEDBACK： 52 dB BASS CONTROL：+17 dB － 15 dB at $50 \mathrm{c} / \mathrm{S}$ ．HUM LEVEL：－75dB HARMONIC DISTORTION at 10 Watts 1,000 c．p．s． $0.1 \%$ ． Complete Kit of parts with full constructional details and point to polnt wiring diagrams． point to polnt wiring diagrams． monthly payments $31 / 1$（Totai £18．3．9）．Or In Teak veneer housing as monthy payments illustrated． 19 Gns UPPLIED BY LEADING BRITISH MANUFACTURFRS． <br> BRADFORD <br> $$
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$$ <br> BRISTOL <br> Lower Castlo St．（Hall．day Wadi）Tol 2209 <br> BIRMINGHAM <br> $3^{30}$ <br> ..... S．t．Western $n$ ．ratato op．Snow Hil <br> DERBY 26 osmaston Red The Sosot（hall－dyy Wed．）Tol． 41361dARLINGTON <br> 18 Priststate（Halt－day Wedi．）Tel．880a3 <br> EmanatTues．，Tel． 3232 2．1512 <br> HI－FI CENTRES LTD． MAIL ORDERS TO： 102 Henconner Lane，Bramley， Leeds 13．No C．O．D．under £1．Terms C．W．O．or C．O．D． Postage 4／6 extra under £2． $5 / 9$ extra under £5．Trade supplled．S．A．E．with enquiries please．HI－FI Catalogue $4 / 6$. <br>  ANY MAKE OF PICK－UP OR MICROPHONE（CTystal，Ceramic， Magnetic，Moving Coll，Ribbon）CURRENTLY AVAILABLE SUPERB SOUND OUTPUT QUALITY CAN BE OBTAINED BY USING WITH FIRST RATE ANCILLARY EQUIPMENT．All required parts， dagrams and detalled instructions 19 ${ }^{\frac{1}{2} \text { Gns．}}$ Unit factory built 27 Gins，or deposit \＆6．2．0 and 9 morrthl 15 Unit ractory buit 27 Gns，or deposit e6．2．0 and 9 monthly peyments 56／3． （Total £31．8．3）．Or in Teak veneer housing 30 Gns．Carr．15／－or Deposit $£ 6.2 .6$ and 9 monthly paymts $64 /-(T o t a l$ $234,18.6)$ ．Send S．A．E．for leafiet <br> > etc, except for Ganglig and Balance Control, apply also to Super 15 THESE HNTTS ARE EMINENTLY SUITABLE FOR USE WTTH etc，except for Ganglig and Balance Control，apply also to Super 15， etc，except for Ganglig and Balance Control，apply also to Super 15， <br> $\qquad$ <br> 73 Dale St．（Half－day Wed．）Tel．CENtral 3573 LIVERPDOL 238 EdgwareRoad，W2（Half－day Thurs．）Tel．PAD 1629 LONDON 98 High Holborn，WC1，Tel．HOL 9874 （Half－day Sat．） 60A Oldham Street（Half－day Wed．）MANCMESTER   （Hall－day Wed．）Tel． 21469 <br> 13 Exchange Street（Castle Market Bldgs．）SHEFFIELD

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 $\star$ Performance comparable with units costing
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 DEMONSTRATIONS AT ALL BRANCHES
R.S.C. STEREO/20 HI-FI AMPLIFIER PRovivi
 0/14 WATT ULTRA LINEARPUSH-PULLOUTPUT ON EACH CHANNEL. SUITABLE FOR "MIKE".
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Frequency response $\pm 3 \mathrm{~dB} 30-20,000 \mathrm{c} / \mathrm{s}$. Sectionally equency response $\pm 3 \mathrm{~dB} 30-20,000 \mathrm{c} / \mathrm{s}$. Sectionally
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Sensitivity 12 millivolts so that any kind of Microphone or Pick-up is suitable. Designed for Clubs, Schools, Theatres. Dance lialls or Outdoor Functions, etc. For use with Electronic Organ, Guitar, String Two inputs with associated volume controls so that two separate inputs such as Gram and "Mike'" can be mixed. $200-250 \mathrm{v}$. $50 \mathrm{c} / \mathrm{s}$ A.C. mains. For 3 and 15 ohm speakers. Complete kit of parts with point-to- 13 Gns.
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TORISED VERSION O TORISED VERSION Of above complete kit 9 Gins
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## A.S.C. BASS-REGENT

 50 WATT AMPLIFIER An exception-
ally powertul ally powertul
high quality high quality all-purpose
anit for lead,
riythm rhythm, bass ists, gram * Two extra heavy ducy Loudspeakers

* Four Jack inputs and two Volume controls for simultaneous use of up to four pick-ups or "mikes"
Bass and Treble controls. Bass and Treble controls.
$49 \frac{1}{2}$ Gins. Carr. $30 /-$ o ${ }^{2}$ monthly payments and 9 monthly payments of
$\mathbf{2 5 . 1 0 . 1 0}$. (Total 55 gns.) 25.10 .10 . (Total 55 gns.) Send S.A.E. for leaffet. 25w Spkr. 291 gns, G15 inc.
121n. 20w Spkr. 194 gns . R.S.C. BATTERY/MAINS CONVERSION UNITS
 Type BMI An all-dry battery eli minator 2in. approx. Completely replaces bat teries supplying 15 v . and 90 y . where A.C. mains 200 250 v . $50 \mathrm{c} / \mathrm{s}$ is available. complete kit with diagram
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FULLY GUARANTEED. Interleaved and ImpregMIDGET CLAMPED TYPE $2 \frac{5}{3} \times 2 \frac{3}{1} \times 2 \frac{1}{1} \mathrm{in}$. MIDGET CLAMPED
$250 \mathrm{v} ., 60 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .2 \mathrm{a}$. FULLY 8HROUDED UPRIGHT MOUNTING $250-0 \cdot 250 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{~s}, 1,0 \cdot 5 \cdot 6 \cdot 3 \mathrm{v}, 3 \mathrm{a}$ $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6-3 \mathrm{v} .4 \mathrm{a}, 0-5-6-3 \mathrm{v} .3 \mathrm{a}$ $300-0-300 \mathrm{v} .130 \mathrm{~mA}, 6-3 \mathrm{v} .4 \mathrm{a}$
For M11lard 510 Amplifer
$350-0 \cdot 350 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a} ., 0 \cdot 5-6 \cdot 3 \mathrm{v} .3 \mathrm{a}$ $350 \cdot 0-350 \mathrm{v}, 150 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}$
$425-0.425 \mathrm{v}, ~$
400 mA $425-0-425 \mathrm{v} .200 \mathrm{~mA}$,
$425-0.425 \mathrm{v} .200 \mathrm{~mA}, 6$
$450 \cdot 0-450 \mathrm{v} \cdot 250 \mathrm{~mA}$
 TOP SHROUDED DROP-TEROUGH TYPE $250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .3-5 \mathrm{~s}$.
$250-0.250 \mathrm{v} .100 \mathrm{~mA}$,
$350.0-350 \mathrm{v} .80 \mathrm{~mA}$.
$200-0 \cdot 250 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}, 0-5-6 \cdot 3 \mathrm{v} .3 \mathrm{~B}$ $300-0-300 \mathrm{v}, 100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v}-4 \mathrm{a} .0 .5 \cdot 6 \cdot 3 \mathrm{v} .3 \mathrm{z}$ Suitable for Mullard 510 Amplifler . $350-6.350 \mathrm{v} .100 \mathrm{~mA}, 6 \cdot 3 \mathrm{v}-4 \mathrm{a} ., 0 \cdot 5 \cdot 6 \cdot 3 \mathrm{v} .3 \mathrm{a}$
$350-0.350 \mathrm{v}, 150 \mathrm{~mA}, 6 \cdot 3 \mathrm{v} .4 \mathrm{a}, 0-5-6.3 \mathrm{v} .3 \mathrm{a}$ FILAMENT or TRANSIETOR POWER PACK TYpes
 $0-9 \cdot 18 \mathrm{v}$. $14 \mathrm{~s} .17 / 8 ; 0-12-25-42 \mathrm{v} .2 \mathrm{a} .29 / 9$. CHARGER TRANSFORMERS 0-9.15v. 1 ta. 14/11分a. 17/9; 3a. 19/11; 5a. 23/9; 6a. 27/9; 8a. 33/8
AUTO (Step UP/step DOWN) TRANSFORMER AUTO (Step UP/step DOWN) TRANSFORMERS $\begin{array}{lll}0-110 / 120 \mathrm{v} .200-230-250 \mathrm{v} . & 50-80 \text { watts } & 15 / 9 \\ 150 \text { watts, } 29 / 11 ; 250 \text { watts } 48 / 9 ; 500 \text { watts } & 99 / 9\end{array}$ OUTPUT TRANSPORMERS
Standard Pentode $5,000 \Omega$ or $7,000 \Omega$ to $3 \Omega$.
Push-Pill 8 watta EL84 to 30 or
Push-Pull 8 watta EL84 to $3 \Omega$ or $15 \Omega$.
Push-Pull 10 watts $6 V 6$ ECL86 to $3,5,8$ or
Push-Pull 10 watts 6V6 ECL86 to 3, 5,8 or 15 12/8 Push-Pul Els 84 to 3 or $15 \Omega 10-12$ wiatt Push-Pull $15 \cdot 18$ watte, sectionally wound 6 L 6 KT66, ete., for 3 or $16 \Omega$ Pound EL34 watt high quality sectionally SMOOTHING CHOKES
$150 \mathrm{~mA}, 7 \cdot 10 \mathrm{H}, 250 \Omega 12 / 9 ; 100 \mathrm{~mA}, 10 \mathrm{H}, 200010 / 8$ $80 \mathrm{~mA}, 10 \mathrm{H}, 350 \Omega 8 / 9 ; 60 \mathrm{~mA}, 10 \mathrm{H}, 400 \Omega 4 / 11$.
${ }^{21 / 9}{ }_{23 / 9}$

## R.S.C. COLUMN SPEAKERS tone Rexine/Vynair, ideal for vocalists two- and vor tone Rexine/Vynair, ideal for vocalists and Public Address. 15 ohm matching. Type C $48,25-1$ 30 watts. Fitted four 8 in . high flux 7 watt speakers. HI•FI CENTREs LTD. Overall size approx. $42 \times 10 \times 5 \mathrm{in}$. Or deposit $65 / \mathrm{F} / 5 \mathrm{Gns}$ Type C412, 40 watts. Fitted four $12 \mathrm{in} .12,000$ ine 10 watt 22 Gns. Or Deposit £3.13.0 and 9 monthly payments of $50 /$ - (Total £26.3.0). <br> 30 WATT HI-FI AMPLIFIER

Aor 2 Input, 2 volume control $\mathrm{Hi}-\mathrm{Fi}$ unit with Latest valves. Strong Rexine . Peak rating 60 watts. handles. Attractive black/gold covered cabinet with ndicator. For $200-250 v$. A.C. mains. 9 1acia. Neon or 3 or 15 ohm speakers. Send S.A.E. 18 Gns. Carr.


12in. HIGH QUALITY LOUDSPEAKERS In Teak reneered or Rexine covered Cabinets 10 Watt 12,000 lines 5 Gns. 20 Watt 12,000 lines 8 Gns.
3 or 15 ohms
LOUDSPEAKERS $L$ Liminitad numbar of heay buty
ohms impedance. Brand new. guaranteed. Terms available over $£ 8$



$18^{\prime \prime} 100$ Watt PROST handing. Guarr 19 Gns
R.S.C. $4 / 5$ watt

## A5 HIGH GAIN AMPLIFIER

A highly-sensitive 4-valve quality amplifier for the mic P.U. heads and most "mikes". Separate Bass and Treble controls. Hum level 71 dB down. Negative Feedback 15 dB . For A.C. mains $200-250 \mathrm{v}$. Speaker output \&4.17.9 3 ohms. Complete Kit with point-to-
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## CLEARANCE LINES

HIGH QUALITY $8^{\prime \prime} \times 3^{\prime \prime}$ LOUDSPEAKERS $\begin{gathered}10000 \text { Gauss } \\ 3 \text { ohm only } \\ \text { 12 }\end{gathered} \mathbf{D}^{2}$ EXTENSION 'SPEAKERS 29/9

EMI PLAYER T/TABLES
Cabinet size $12 \times 8$ x $5 i n$.
attractive grey lizard skin finish.
With P.U. 4 speed. T/D cartridge Fitted high flux 6tin. 5w. 3 ohm

Mono 79
89/9

## PHONE AMPLIFIERS



## 1 WATT TRANSISTOR AMPLIFIERS

for 3 3.5 ohm $39 / 11$

## PRINTED CIRCUIT KITS

$\qquad$ or making printed circuits.
14/11

## J.B. VHF/FM DIAL \& DRIVE ASSEMBLIES

## TAPE RECORD/PLAYBACK AMPLIFIERS


MINI-8 HI-FI LOUDSPEAKER UNITS


HEAVY DUTY 15in. 40 WATT LOUDSPEAKERS
14 Gns STEREO/TEN HIGH QUALITY AMPLIFIER

5 watts high quality output on each channel.
Sensitivity 50 millivolts. Suitable all crystal or ceramic stereo heads. Ganged Bass and Treble

0.000 .0

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OAZ20012j－ \& OAZ208 \& $6 / 6$ <br>
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$2.4 / 2.7 / 3 / 3.9 / 4.3 / 13 / 16 / 18 /$ $20 / 30 / 33$ volt， $5 /-$ esch $Z$ series．All voltages from $1.5 \mathrm{w} .4 / \mathrm{e} \mathrm{ea} .7 \mathrm{w} .5 /-$ each


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 ance $200 / 250 \mathrm{~V}$ A．C．
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High quslity instrument with 28 ranges．D．C．volts $1.5-1,500 v$ ．A．C．Volts $1 \cdot 5-$
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ing 120
$\mathrm{Kc} / \mathrm{s}-260$
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6 Directly callibrated t． tenuator，Operation $200 / 240 \mathrm{~V}$ ．A．C．
Brand new with in－ P．\＆$\quad$ P． 7 for details．
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 A．F．SINE WAVE
$20-200,000$
C／8
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VOLTMETER | 30，000 c／s． | $0 / P$ | 10 meg，input 10 ranges： |  |
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 TF $100 \mathrm{Kc} / \mathrm{s}-300 \mathrm{P} \quad 1-2 \mathrm{Mc} / \mathrm{s}$ ．Decibeis -40 to attenuation int／ext．modalation．Incorpor－new complete with leads put and \％mod．on E F．220／240V．A．C．tion 230 V ．A．C． 217.10 .0 ． e30．0．0．Cart．7／6．
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pF－2，000 mFd 2 ohms 200 meg．
ohms．Also checks impedance，turns
ratio，insulation $200 / 250 \mathrm{~V}$.
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／6．R． $220 / 240 \mathrm{~V}$

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meter protection． $0 / \cdot 5$ $2 \cdot 5 / 10 / 50 / 250 / 500 /$
1.000 V D．C． $0 / 3 / 10 / 50 / 2$ $250 / 500 / 1,000 \mathrm{~V}$ A．C． 01 $10 / 100 \mu \mathrm{~L} / 10 / 100$／ $500 \mathrm{~mA} / 2 \cdot 5 / 10 \mathrm{~A}, 0$ ）
$1 \mathrm{~K} / 10 \mathrm{~K} / 100 \mathrm{~K} / 10 \mathrm{M} /$ $10 \mathrm{MO}-10$ to $49 \cdot 4 \mathrm{~dB}$.
218.18 .0 ．P． P ． LA AMEMLE 57 Range nper 50 E ／／volt Manti－ -1000 V ．A．C．volts 125 mV -1000 V ．D．C．volte current $25 \mu \mathrm{~A}$－10 amp．Ohms $0-$ $0 \mathrm{meg} \Omega \cdot \mathrm{dB}-20$ to +81 18．10．0．Carr


PROFESSION－
AL 20,000 opv AL 80,000 opv
LAB．TYPE Maltiteater． Automatic overlcud pro－ salc．Rangea D．C．and A．C Current： $0 / 20 \mathrm{~K}$

$110 / 50 / 250 / 500 / 1,000$ volta $00 \mathrm{~K}, 2 \mathrm{megohm}$ ．Decivels
 P．\＆P． $2 / 6$ ． 22 dB ． 85.19 .6 ． MODEL TE－70．$\quad 30,000$ ．P．V．0／3／15／60／300／600
 $3 / 30 / 300 \mathrm{~mA} . \quad 0 / 16 \mathrm{~K} / 160$ $\mathrm{K} / 1 \cdot 6 \mathrm{M} / 16 \mathrm{Meg}$ ． 8． 50.0 P．\＆P． $3 /-$
$\begin{array}{lr}\text { MODFL } & \text { 250J．} \quad 80.00 \\ 0 . P . V . & 0 / 10 / 50 / 500 /\end{array}$ $\begin{array}{ll}\text { O．P．V．} & 0 / 10 / 50 / 5001 \\ 2.500 \mathrm{v} . & \text { D．C．} \\ 0 / 10 / 501\end{array}$ $500 / 2,500 \mathrm{v} . \quad$ A．C．$\quad 0 / 10 / 50 / 2$ Meg． $\mathrm{a}^{\circ}$ ． $0 / 250 \mathrm{~mA}$ ． 40／6．P．\＆P．2／6．

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proteotion．fin，full view raeter． 2 colour scale． 0 $2 \cdot 5 / 10 / 250 / 1,000 / 5,000 \mathrm{v}$ ． A．C．O／25／12－5／10／50／250 1，000／5，000v．D．C． $0 /$ $50 \mu \mathrm{~A} / 110 / 100 / 500 \mathrm{~mA} /$
10 amp ．D．C．
$02 \mathrm{~K} /$ $200 \mathrm{~K} / 20 \mathrm{MEG} . \mathrm{OHM}$ ． 815．P．\＆P． $5 /$

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Built－in ineter protection $0 / 3 / 12 / 60 / 120 / 300$ $120 / 300 / 600 \mathrm{v}$ ． 20／300／600v． 12 amp． $0 / 2 \mathrm{~K} / 300 \mathrm{~mA}$ 12 amp． $0 / 2 \mathrm{~K} / 200 \mathrm{~K} / 2 \mathrm{M}$
200 m 0.
20 to +17 BB \＆12．10．0．P．\＆P．3／6．


## MODEL



MODFL
APCale，built
bukn $2 / V$ geale，built－in meter pro－ tection． $0 / \cdot 3 / 3 / 12 / 60 / 120 /$ $300 / 600 / 1,200 \mathrm{v}$ D．C． $0 / 6 /$ $30 / 120 / 300 / 600 / 1,200 \mathrm{v}$ ．A．C．
$0 / 30 \mu \mathrm{~A} / 6 / 60 / 300 \mathrm{~mA} / 12$
Amp． $0 / 10 \mathrm{~K} / 1 \mathrm{~m} / 10 \mathrm{M} / 100$ $\begin{array}{ll}\text { Amp．} & 0 / 10 \mathrm{~K} / 1 \mathrm{M} / 10 \mathrm{M} / 100 \\ \mathrm{M} \Omega . \\ 88.10 .0 . & \text { P．\＆P．} 3 / 6 .\end{array}{ }^{+17 \mathrm{~dB}}$

MODEL TEL－12 20，000 O．P．V．

$0 / 0 \cdot 6 / 6 / 30 / 120 / 600 / 1.200 /$ | $3,000 / 6,000 \mathrm{v}$ ．D．C． $0 / 6 / 30 /$ |
| :--- | $\begin{array}{ll}3,000 / 6,000 v . ~ D . C . ~ & 0 / 6 / 30 / \\ 120 / 600 / 1,200 v . ~ A . C . ~ \\ 0 / 60 \mu A\end{array}$ $\begin{array}{ll}120 / 600 / 1,200 \nabla & \text { A．C．} 0 / 60 \mu \mathrm{LA} \\ 0 / 6 \mathrm{~K} / 600 \mathrm{~K} / \mathrm{G}\end{array}$




MODEL TE－80．20，000 O．P．V．0／10／50／100／500／ $1,000 \mathrm{v}$ ．A．C． $0 / 5 / 25 / 50 / 250 /$ $500 / 1,000 \mathrm{v}$ ：D．C． $0-60 \mu \mathrm{~A}$ ， $\begin{array}{ll}5 / 50 / 500 \mathrm{~mA} . & 0 / 6 \mathrm{~K} / 60 \mathrm{~K} / 600 \\ \mathrm{~K} / 6 \mathrm{meg} .84 .17 .6 . ~ P . ~ \& ~ P . ~ 3 /-. ~\end{array}$

MODEL PT－34． 1,000 OOV．0／10／50／250／ D．C． $0 / 1 / 100 / 500 \mathrm{~mA}$ 1．C．C． $0 / 100 \mathrm{~K} \Omega$ ． $89 / 6$ ．
P．\＆ $\mathbf{P} .1 / 6$ ．


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 19 transistors， 8 diodes， 1 HF music power，30W at $8 \Omega$ ．Response $30-20,000 \pm \frac{2 \mathrm{~dB}}{\mathrm{st}} 1 \mathrm{~W}$ ．Dis－
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## TOPIC OF THE MONTH

## A Matter of Mode

SINCE time immemorial, or at least since your editor has been connected with radio (which some cynics claim is synonymous), the theme of "CW versus Phone" has simmered gently, periodically coming to the boil. Much of the comment is belligerently partisan and a lot of snide remarks are directed at the key bashers (mainly by those who have yet to pass the Morse Test).

It is contended that CW is an outmoded form of communication, an anachronism in these modern swinging times. Nevertheless, the detractors would be surprised, if only they could read Morse, at the large number of amateurs still wielding a key.

The object of amateur radio is communication and since CW is one mode it is not unreasonable to expect licensees to be able to use it. The Morse Test is retained by every country (at least for h.f. band usage) as a necessary hurdle to overcome. One might as well argue that telephony operators should not be expected to pass the RAE!

On the practical side, CW is by far the most efficient method of communication and occupies far less bandwidth than telephony (even SSB). When a band is "dead" to the telephony-only operator, the key pounders are usually able to keep going. During periods of poor conditions, it is often the only way to get through and the operator unable to switch to this mode is really only half an amateur.

With telephony, a contact between two stations in different countries requires that one or another of the operators must know both of the languages concerned. Most English amateurs are not linguists and it is arrogant to expect all foreign amateurs to know our language. This barrier does not exist to such an extent on CW because the international codes and abbreviations permit contacts which at least provide the basic information required at each end.

On the other hand, telephony can have many advantages and we would be the first to admit this. Let us, therefore, kill off the concept of "CW versus Phone" and substitute "CW and Phone". Both have their place in amateur radio and all stations should be equipped to handle both modes of communication.
W. N. STEVENS—Editor.

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ON NOVEMBER 8th

[^1]JET KING MINI BLOW TORCH


The Walter Kidde Company Limited has introduced, in the U.K., the Jet King blow torch and soldering tool.

This pocket-sized unit will allow you to solder or braze with precision in "tight spots" inaccessible to the usual tool. It is suitable for radio and TV work, repairs around the house, model making, etc.

The Jet King is completely portable and self-contained and gives approximately 45 minutes continuous burning with each butane charger. It is available in all leading departmental and hardware stores and garages. The cost is 35s. complete with two chargers. Refill chargers are readily available in packs of two for $6 s .6 d$.

## SWITCH CATALOGUE

Arcolectric Switches Ltd, Central Avenue, West Molesey, Surrey, England, announce their Catalogue No. 137. It describes their complete range of products and replaces all previous catalogues. It is divided into seven sections.

Section D covers transformer signal lamps. The remaining sections covering switches, neon indicators and signal lampholders are generally as before, but include some new styles and types.


## belling-LEE TERMINALS

A new series of miniature terminals, with a current rating of 10 A and a breakdown voltage greater than 4 kV d.c., is announced by Belling \& Lee Limited.

Known as type L1726, these new terminals are only a quarter the size but have all the features of the standard style which they resemble in appearance. These include a captive head available with a choice of six standard colours, a socket in the top for plugging in auxiliary miniature connections to extra monitoring equipment, and a cross-hole in the clamping gap which accepts wires up to 1.9 mm diameter ( 0.075 in., i.e. $15 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. solid wires or $40 / \cdot 0076 \mathrm{in}$. stranded conductors).

## RADIO AT THE EDINBURGH INTERNATIONAL FESTIVAL

Lothians Radio Society, GB3EIF, operated a demonstration Amateur Radio Station during the period of the Edinburgh International Festival 1968 (17th August to 7th September).

The station was located at the Heriot Watt University, Mountbatten Building, Grassmarket, Edinburgh 1, and was operational on the bands $80-10 \mathrm{~m}$ using s.s.b. equipment. An additional S.tation was operational on 2 m using phone/c.w.

A special Q.S.L. card was issued to confirm all Q.S.O's and there was an exhibition of equipment built by members of the society.

ELECTRONIC LABORATORY KIT

A.P.I. have introduced a new "Lektrokit" No. 6 electronic construction kit for laboratory and educational use. It consists of a bench rack with two chassis assemblies (see picture) on which discrete components and integrated circuits can be mounted and wired. A front panel of grey enamelled aluminium alloy is also provided for indicator lamps, meters, switches, and other controls.

The base tray can be used to carry power supplies or other auxiliaries, and there is space for an additional chassis assembly or front panel if required.

This kit enables laboratory experimental circuits to be neatly stacked vertically instead of the usual sprawl across the bench, and allows immediate "shelving" of a rig if a need arises to clear the bench space for other work.

The price of the Lektrokit No. 6 is $£ 710$ s. and further details are available from the manufacturers, A.P.T. Electronic Industries Ltd., of Byfleet, Surrey.

## STEREOPHONY FOR THE NORTH

From 10 August the stereo programmes on Radio 3 were extended to a large part of Northern England. The transmissions are on $91.5 \mathrm{Mc} / \mathrm{s}$ from Holme Moss.

## MICRO Hi-Fi



Sinclair Radionics have introduced the world's first monolithic integrated circuit hi-fi power and preamplifier. The complete unit, known as the IC-10, measures only $1 \times 0.4 \times 0.2 \mathrm{in}$. and the circuit which this package contains is a chip of silicon just one-twentieth of an inch square by one-hundredth of an inch thick which contains 13 transistors (including two power types), 3 diodes and 18 resistors.

The circuit is complete in itself requiring only the addition of tone and volume controls and a power supply which may be battery or mains. Output power is 10 watts peak and 5 watts r.m.s. continuous. Distortion is less than $1 \%$ at full output and frequency response is $5 \mathrm{c} / \mathrm{s}$ to $100 \mathrm{kc} / \mathrm{s} \pm 1 \mathrm{~dB}$.

The preamp section contains 3 transistors with cut-off frequencies in excess of $500 \mathrm{Mc} / \mathrm{s}$ so that it may be used as an r.f. or i.f. amplifier in addition to its normal audio application. The power amplifier is a 10 transistor circuit with a class $A B$ output stage and closely defined quiescent current which is independent of temperature because of the extremely close thermal coupling between the output transistors and the bias diodes. The gain of the entire unit is accurately defined by built-in negative feedback loops and the circuit may be operated down to d.c. for special applications because direct coupling is employed throughout.

The amplifier will operate on any power source providing between 8 and 18 volts and is suitable for driving speakers of any impedance between 3 and 15 ohms.

The $1 \mathrm{C}-10$ is guaranteed for 5 years and costs only 59 s . 6 d . complete with a comprehensive instruction manual.

## ROBERTS "IC" PORTABLE

Roberts Radio Co. Ltd. have launched their 2-band portable receiver type R/C.1. It is the first British radio receiver to employ a tiny monolithic integrated circuit. This device embodies most of the active components between the aerial and the output stage and a crystal filter is used for selectivity.

BOOKLET ABOUT THE RRE
The Royal Radar Establishment at Malvern, Worcs. is the largest centre for electronics research and development in Britain. It has now produced a booklet entitled "RRE Activities Guide" which gives an overall view of the programme of work at Malvern, from basic physics research to the development of advanced electronic equipment.

Copies of the booklet are available free from the Industrial Applications Unit, Royal Radar Establishment, Ministry of Technology, St. Andrews Road, Great Malvern, Worcs.

BRITAIN HELPS U.S. AIRLINE
Pan American World Airways is using a Mufax Courier facsimile communications system, supplied by the Muirhead Group of Beckenham, for sending facsimile copies of messages from nine transmitters in widely dispersed departments at Heathrow to the airline's European Telecommunications Centre on the south perimeter road. From there, the messages are transmitted over Pan Am's private teleprinter network to its stations throughout the world.

## SOLDERING SAFETY



Weller have now introduced a range of simple, safe bench holders for both their industrial (TCP-1, W60D, W100D) and their consumer soldering irons (SP25D, SP40D, SP80D).

Prices range from 16s 0d. and, in addition to the holder and base, each kit includes a sponge for the easy cleaning of the soldering iron tip.

COLOUR CODING OF CABLES AND CORDS It was announced recently that the Government have decided to make regulations under the Consumer Protection Act requiring the core colours of three-core flexible cords fitted to domestic electrical appliances, when offered for sale in Great Britain, to comply with the following international coding recently agreed by most of the countries of Europe:

Green-and-yellow striped core, Earth; Brown core, Live; Light blue core, Neutral.

## HI-FI IN DUSSELDORF

The high fidelity audio fair "Hi-Fi 68" opened at Dusseldorf, West Germany on August 30 to September 3. One hundred and twenty three companies from nine different countries including Japan, America and the United Kingdom took part. Demonstrations of Hi -Fi were given in sixty specially sound-proofed cabinets.

## PICO AMP DIODE

SGS-Fairchild have just announced their BAW30 miniature (TO 46) diode. It has a reverse current of only $10 p \mathrm{~A}$ ( $V_{R}=10 \mathrm{~V}$ )." Capable of handling forward currents up to 100 mA , this diode has a power dissipation rating or 125 mW at $25^{\circ} \mathrm{C}$. Other ratings include a breakdown voltage greater than 35 V at a reverse current of $1 \mu \mathrm{~A}$, forward voltage drop of less than 1 V at 10 mA forward current and a capacitance of less than $1 \cdot 3 p F$ at zero reverse voltage. Fuller details from SGS-Fairchild, Planar House, Walton St., Aylesbury, Bucks.

# Transistor Gram Amplifiers 

## R.F. GRAHAM

TRANSISTOR record playing equipment is easily made, and the amplifier and speaker may be separate items or contained in the same case as the turntable mechanism. Economical gram motor units to run from a 9 V or similar dry battery are easily obtained, and the player is then portable and needs no mains supply. It is also quite feasible tc employ a mains-driven turntable assembly, feeding its output to a transistor amplifier.

## Low Volume Circuits

For single speaker reproduction of records or headphone listening to instructional records etc. a very simple amplifier is adequate. The circuit shown in Fig. 1 easily gives enough volume for this, with a $33 / 45$ r.p.m. crystal pickup. VR1 is the volume control.

The OC72 operating conditions are set by the biasing components for a collector current of about


Fig, 1: Circuit of a two-stage amplifier.
15-20 mA. TI can be a transistor type output transformer as used for push-pull output stages, with the centre-tap connection simply ignored. Alternatively a $70 \Omega$ or similar speaker may be connected from the collector to the negative supply line. Personal phones or headsets can also be connected in this way. Low-impedance speakers (such as $2 / 3 \Omega$ units) must be connected to T1 secondary.

Cl may be changed in value to modify tone. If the motor is for 6 V or 9 V , this may also be used for the amplifier.

## Push-Pull Output

For many purposes an output of about 500 mW is enough and can be obtained by using two OC81 or equivalent transistors. Figure 2 shows a driver and push-pull circuit of this type. For a smaller output
two OC72 or similar transistors may be used, with an OC71 as driver. With OC72 output transistors R1 is $2.7 \mathrm{k} \Omega$ and R2 $100 \Omega$. A two-stage amplifier of the type shown in Fig. 2 can give excellent volume with a fairly large output type pickup. Many ordinary crystal pickup units are satisfactory.

The volume may well be adequate with Tl omitted and the pickup volume control connected as in Fig. 3. The $100 \mu \mathrm{~F}$ capacitor is then not required, and the base of the driver is taken direct to the junction of the $47 \mathrm{k} \Omega$ and $12 \mathrm{k} \Omega$ bias resistors.

## Pickup Matching

A crystal pickup is intended to work into a highimpedance load, while the input impedance of the first transistor in circuits such as Figs. 1 and 2 is quite low. Impedance matching may be arranged by a transformer, e.g. T1 in Fig. 2, or a resistor, R1 in Fig. 3. A transformer transfers the pickup output to the transistor base with greater efficiency, allowing maximum volume with a given pickup and number of transistors. The ratio of Tl is not too critical, but it needs to have a high primary inductance.

In Fig. 3 R1 presents a high-impedance load at any setting of the volume control but also causes a loss of signal strength. R1 may be reduced in value if some drop in quality is not very important and this increases volume. With the arrangement shown in Fig. 3 an additional stage of amplification is usually needed. However, the few components and transistor needed may easily cost less than a matching transformer while the chances of picking up interference such as motor noise by the transformer are avoided. So this technique is often preferred unless a suítable transformer happens to be to hand.

With all circuits the pickup can be permanently


Fig 2: Driver and push-pull amplifier circuitry.


Fig. 3 (above): Matching circuit for highimpedance pickup.

Fig. 4 (right): One-watt amplifier circuit with the impedance matching circuit of Figure 3 incorporated. A crystal type pickup may be used that has an output of approximately 300 mV .

connected or have a coaxial plug to fit a socket on the amplifier. The latter allows the same amplifier equipment to have other uses, such as with a radio tuner. The outer brading of the pickup lead is the "earth" or battery positive connection.

## One-Watt Amplifier

For very many purposes an output of up to 1 watt is adequate and does not impose too much drain on a dry battery supply. Two OC81 or equivalent transistors will provide this output using the circuit in Fig. 4. A high-gain preamplifier stage follows the pickup, boosting signal strength for the driver stage. A crystal pickup with a nominal rated output of about 300 mV is suitable.

For operation at maximum volume each output transistor should be in a clip which is bolted to a heatsink. The latter can be made from pieces of $16 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. aluminium about $3 \mathrm{in} . \times 2 \mathrm{in}$. A flange is bent on the aluminium so that each plate stands vertically.

The same circuit can be used with outputs less than 500 mW without heatsinks. R1 can then be changed to $4 \cdot 7 \Omega$.
Negative feedback can be omitted by not including R2, R3 and C1. Or R2 may be omitted or changed in value and C1 and R3 included. If R2 is present Cl and R3 should not be omitted. Feedback improves quality and if volume permits R2 may be reduced to $470 \mathrm{k} \Omega$.

The OC71 preamplifier stage in Fig. 4 can be added to the circuit shown in Fig. 2 or other circuits where the pickup output or method of matching results in insufficient signal to drive the output stage. In these circumstances good results should be secured with almost any ordinary driver plus push-pull output circuit.

## Complementary Push-Pull

A transformerless circuit operating directly into a $15 \Omega$ speaker is shown in Fig. 5 and is very satisfactory for record reproduction. A reasonably large
speaker unit is best and can give excellent reproduction and more than adequate volume for domestic purposes. The output transistors may be damaged by using a speaker of wrong impedance. The first stage is a high-gain amplifier, coupled by C3 to the driver stage which is directly coupled to the output transistors. The NKT279.SI transistor provides compensation for the effect of heat on the operation of the driver and output transistors, the $50 \Omega$ resistor being initially adjusted so that the output transistors draw 1.5 mA to 2 mA with no signal. Current peaks to $20-30 \mathrm{~mA}$ at good volume.

## Volume Control

The volume control may be mounted at any convenient position in the cabinet. Screened leads run from the pickup to the volume control and from the latter to the amplifier. These leads can be soldered, with a little extra length so that the amplifier can be taken from the cabinet if necessary. Should an amplifier be constructed as a separate unit for use


Fig. 5: Transformerless amplifier operating directly into a $15 \Omega$ loudspeaker.
with a pickup, radio tuner, etc. the volume control is generally fitted to the amplifier case. A coaxial or other connector allows any equipment to be plugged into the amplifier input.


Fig. 6: Utilising the gram auto-switch to control the amplifier.
Fig. 7: Component layout and wiring of the amplifier shown in Figure 2.


## On/Off Switch

For a separate amplifier the on/off switch may be incorporated with the volume control and placed in one battery lead. If the amplifier is installed permanently in the player the automatic switching device normally found for the motor can be used to switch on the amplifier also. Then when the record is finished both motor and amplifier are automatically switched off. Figure 6 shows the circuit for this. Polarity to motor and amplifier must be correct and errors will be avoided by fitting the proper type of battery connectors.

## Battery and Motor

The current drawn by the motor depends on its voltage and other factors, but about 15 mA from a 9 V supply should be expected. This is easily within the capacity of a PP9 or similar battery, or a supply made up from non-miniature torch cells or batteries.

With push-pull output stages the current peaks depend largely on the volume but will generally be around $20-40 \mathrm{~mA}$. However battery drain falls to a much lower level when strong audio signals are not present.

## Interference

With normal care motor interference need not be picked up by the volume control or associated input circuits. Leads here should be screened, dressed away from the motor, and the volume control case earthed to battery positive. The metal parts of the turntable unit should be similarly earthed.

Interference from the motor may also reach the amplifier through the common battery circuit. A $100 \mu \mathrm{~F}$ or similar large value capacitor across the battery supply points at the amplifier (as C8 in Fig. 5) helps avoid this. Earlier stages of the amplifier are generally fed from a negative line which has a series resistor and capacitor (as in all circuits except Fig. 1) so do not cause trouble.
Should any form of higher frequency interference arise a capacitor effective at high frequency can be shunted across the $100 \mu$ F electrolytic. Small 200pF disc ceramic capacitors can be tried or an $0.5 \mu \mathrm{~F}$ paper or similar component. Small suppressor chokes may also be included in the leads to the motor only. These can be made of a few score turns of 26 s.w.g. or similar wire. Some motor units have ferrite beads on the motor supply leads to give a similar effect.
If motor interference should arise with some particular set of equipment a simple test is to use a separate battery temporarily for the amplifier. If the interference ceases it is caused by the common battery circuit. If it continues it is picked up by the volume control or other input circuits.

## Pickup

Most turntable units have a suitable crystal pickup fitted. The turntable assembly and pickup must of course be chosen to suit the speeds required. Popular crystal pickups of this kind supply quite


Upper photograph shows the 3-transistor amplifier (Figure 2) and below it is shown the amplifier circuit depicted in Figure 5.
a large output and changing from such a unit to a low-output or high-quality transcription pickup will often make it necessary to add an extra transistor preamplifier stage.

## Amplifier Construction

The amplifiers can be assembled on $1 / 16$ in. thick paxolin sheet with the components on one side and the wiring on the reverse. Figure 7 is a suitable layout for the circuit in Fig. 2. The ends of the resistors and other components pass through small holes. The panel is then turned over and component interconnections made with 26 s.w.g. tinned copper wire
with 1 mm . sleeving as necessary. Snip off excess resistor and other leads. Transistors are inserted with leads through the holes shown.

It is convenient to solder on thin red flex for battery positive, black for battery negative, and twin of some other colour for the speaker. Also a screened lead for input. The amplifier can be fixed at some convenient position in the case by two or three small screws.

Figure 8 shows a suitable layout, with wiring, for the circuit in Fig. 5. R13 is a miniature preset resistor and its slider is set to give the no-signal current previously mentioned.


Fig. 8: Component layout and wiring of the amplifier in Figure 5.

No particular difficulty is likely to be met in making and using small audio amplifiers of the type described in this article. Where two transistors are employed in push-pull (as for example in Fig. 2) these are best obtained as a matched pair. If purchased separately their actual characteristics may not be sufficiently similar. With a directly coupled circuit such as that shown in Fig. 5 it is recommended that $\operatorname{Tr} 2, \operatorname{Tr} 3, \operatorname{Tr} 4$ and $\operatorname{Tr} 5$ are purchased as a matched set.

Where the push-pull operation of an output pair such as that in Fig. 2 is unsatisfactory this can generally be corrected by slightly changing the value of R1 or R2. In commercial equipment one of these resistors may be preset. For example R2 in Fig. 2 may be $50 \Omega$, allowing adjustment for optimum value. Alternatively a preset resistor or fixed and preset components in series might allow R1 to be adjusted a few hundred ohms each side of $2 \cdot 2 \mathrm{k} \Omega$.

## THE G.P.O. LICENSES 100,000 PRIVATE RADIO TRANSMITTERS

Over 100,000 radio transmitters for private two-way communication are now licensed by the Post Office in the U.K.
What began 20 years ago as a modest experiment in "business radio" has now grown into a popular facility with a wide variety of users, ranging from ambulances, taxis and motoring organisations, to doctors, vets, public utilities (electricity, gas, water, etc.), organisers of sporting events and mountain rescue teams.

There are 75,000 land-based transmitters for private mobile radio. Over 60,000 of these are in vehicles, enabling, for example, maintenance vans on their rounds to keep in constant touch with central control points. Over 6,000 are small portable sets, such as those used by road and building contractors. About 7,000 are the base stations controlling mobile networks-fixed transmitters, operated from an office or a control point.
The Post Office recently added to the list a few licences to use radio for training blind people to be more independent, and for teaching deaf children to speak.

Over $\mathbf{2 , 3 0 0}$ more transmitters licensed for private communication around our shores are carried by ships of various types from tankers to yachts. Nearly 100 lighthouses use radio. British aircraft (including gliders) and their ground stations account for nearly $\mathbf{2 . 5 0 0}$ more.

Use of mobile radio by police and fire services brings the total to over 100,000 .

The accommodation of so many transmitters has been possible by using v.h.f. and more recently u.h.f. The war years gave considerable experience in v.h.f. communication and advantage has been taken of this and of subsequent technical developments. Radio frequencies cannot be made but they can be used more and more economically: for example, the frequency space required for one private land mobile radio channel has been progressively reduced so that eight channels can now be fitted into the space which 15 years ago would hold only one. This is a little known revolution achieved with the co-operation of industry, users and the Post Office. Receivers contribute to this economy. They must respond well to the wanted signal while rejecting signals on other frequencies, otherwise they limit the number of transmitters which can be used.
These 100,000 transmitters are not of course the whole story. There are also 16,000 licensed radio amateurs who are qualified to use transmitters in this country. Another 14,000 licence holders use radio to control model vessels, vehicles or aircraft. Seven hundred paging systems are in use, mostly the short-range "bleepers" for calling staff in a hospital or in a factory, etc. And there are nearly 3,000 induction communication systems, most of which are a variant on the paging system. They are used on smaller sites, such as single buildings, by a variety of users.

All these have to be fitted into the frequency spectrum in an orderly way, together with longrange services, broadcasting, navigational aids, satellites, and so on. This is why the Post Office emphasises to those who want to use radio how essential it is to get a licence and to use the right sort of equipment. The licence is not just another piece of paper. It is the result of careful planning, national and international, and it explains what has to be done to fit in with the manv other users.

## Noises on the bands

Regarding Mr. Meachim's letter of last month about strange noises transmitted over the ether by Amateurs, I feel that his local lads have got nothing on my lot! In my area there is an Amateur who puts his one-man band on the air! In fact he is pretty good and it is perfectly legal to do this provided he doesn't infringe the copyright laws. I suppose it is legal to have a $50-$ piece male voice choir on the air provided they all say their callsigns every fifteen minutes and sign their names, callsigns and addresses in the log book.

One day 1 was tuning around on Topband and heard someone calling CQ and he had added an echo effect by means of a tape recorder amplifier and some intricate feedback effect. It made him sound as though he were transmitting from a sewer or the Albert Hall when empty.

When he got a reply to his call, he said that since adding this gimmick the number of QSO's had increased quite appreciably.

I don't really know what comes next. My advice to all Amateurs is to scrap their $G$ callsigns and replace them with 3 W 8 ones and have tape loops going in the background with the sound of exploding shells and bombs etc. This, I am quite sure would increase very considerably the number of QSO's.-J. Creedy (Ramsey, Huntingdon).

## 'G8' fights back

I was surprised to read Mr. E. Mason's letter in the July issue of P.W. regarding beginners' licences. The main text of his letter seemed to run down the newly licensed G8 prefixed Amateurs.

How can we call these people "All and sundry" when the only part of the Amateur exam they have not passed is the Morse Code. I fail to see how this makes them any the less intelligent than someone holding a full licence.
1 suggest Mr. Mason listens to the G8's operating on 2 m as I have. I am sure then he will change his mind because they have brought life and interest to a dead but interesting band.-R. Powell (Reading, Berkshire).

## Those odd hours

Your Topic of the Month reminds me of those days when spare rooms, lean-to-shacks and kitchen tables all helped to give birth to what is now a massive industry.

All of us who buy P.W., in what way do we pursue our hobby? I cast my thoughts back over the past few weeks to see just what I had been doing with those odd hours.

A Sky Queen of 1953 vintage which I rescued from oblivion seemed far too nice a job to be cast aside, so 1 converted it to alltransistor, and now have an excellent transportable for house and garden. In the course of conversion 1 fitted a more efficient loudspeaker. A good loudspeaker in a solidly constructed case makes all the difference to transistor output. This project also showed me that a good frame aerial leaves a ferrite rod aerial far behind.

What to do next? Having some silicon transistors to hand, I decided to try a BBC-1 TV preamp. This went well with ST140, TIS18, BC168 and even 2G417. I was stuck for a coax socket. A B7G valve holder performed the task excellently. A bare wire was threaded through the slots in the pin socket soldering tags, drawn tight until there was a springy contact with the coax plug outer and soldered in position. The central socket was a good fit for the plug pin.

There was nothing for it other than to add another stage. I soldered 30 pF beehive trimmers into the screening as tuning capacitors. With this gadget I was intrigued to find that when on the normal aerial foreign interference laid on a jazz pattern which destroyed the picture, with the preamp in circuit, with careful adjustment of the two trimmers I could obtain a watchable picture and still retain sound. I also found that on certain channels the preamp could be used to tune in on the TV sound the three v.h.f. programmes from Skriaig, Isle of Skye, at excellent volume and quality. One thing I found from these experiments was that by using miniature presets in the bottom half of the base pair of resistors, spot-on transistor operation could be obtained without any noticeable capacity side-effects.

From here it was a short step to
wondering what signals were obtainable at higher frequencies from the tops of our local hills. This meant something portable, and led to the searching of old issues and the construction of various superregenerative receivers, and this project continues at the moment, with indications of success around $205 \mathrm{Mc} / \mathrm{s}$ at 1000 feet.

And that, Sir, was my hobbying for the past four weeks.-R. A. Ball (Ross-shire, Scotland).

## Audio thump

I feel sure that the "thump" Mr. Pinder (Questions Answered, August) complains of is almost certainly due to the output d.c. blocking capacitor charging via the speaker and not the smoothing capacitors.

Assuming the amplifier is a Class B design, fitting a thermistor would be disastrous in that the power supply regulation would be ruined. A $47 \Omega$ resistor would also spoil regulation but to a lesser degree.



The best cure for the thump caused by CI (Fig. a) is to fit CA and CB as in Fig. b. Both CA and CB are equal to half the value of Cl (probably $2000 \mu \mathrm{~F}$ ).

With the configuration in Fig. b, CA and CB charge in series and the thump is removed.

Circuit 'a' could employ n-p-n or p-n-p types of transistor and Cl could be connected to the power line or earth line via the loudspeaker. - C. Boggis (Braintree, Essex).

## Scots ahoy!

If any readers are interested in the formation of an Amateur Radio Group in the Scottish Borders, will they please contact me?-G. Shankie, GM3WIG (8 Ettrick Terrace, Hawick, Roxburghshire).

## Misconceptions

The letters of messrs Smith and Gibbins (Sept.) contain some misconceptions. Old germanium transistors of the type of OC45, XA 102 and others can be persuaded to perform at much higher frequencies than quoted by the makers.
The conservative "alpha cut-off frequency" was used, meaning simply that gain is 3 dB down at that frequency, not a lot, and a long way from "maximum frequency", at which gain is unity. At 5 times cut-off frequency amplification is 17 dB down, a transistor of hfe 100 would still give useful gain at 20 dB down.

Transistors of the same type are never sufficiently alike to make any rule invariable. $\beta 50,6 \mathrm{Mc} / \mathrm{s}$ may imply a manufacturing spread of 0 to $500 \mathrm{\beta kc} / \mathrm{s}$ to $12 \mathrm{Mc} / \mathrm{s}$. Of these the inspectorate may pass say 20 to 100 beta, $3 \mathrm{Mc} / \mathrm{s}$ upwards. In use, a $20 \beta, 3 \mathrm{Mc} / \mathrm{s}$ transistor carries the same type number as a $100 \beta$, $12 \mathrm{Mc} / \mathrm{s}$. Surplus transistors vary even worse. A combination of very high gain and exceptionally suitable base enables some individual transistors to far outperform the original specification based upon an average transistor. And even the average transistor can be stretched in performance if used with knowledge.

The upper frequency limit can be raised by up to $50 \%$ by using the transistor at the optimum current for gain (usually around 1 to 2 mA ), and at the maximum allowable voltage. This is because maximum frequency is a function of both current gain and suitability for frequency.-T. E. Millsom (London, $N .19$ ).

## Class " $B$ " fights back

Regarding Mr. Mason's comments last month (July issue), he seems to be condemning the Class " $B$ " licence rather than the proposed Beginner's Licence.

He does not appear to have given the original conception of the class " $B$ " licence much thought.

The idea of the class " $B$ " licence was to increase the activity on the 70 cm and other u.h.f. bands (which it did admirably). Allowing the " $B$ " licensees to use two metres is simply to increase activity on this
band and thus helping to stop the reduction of amateur bandspace, which is being turned over for commercial use etc. Therefore by condemning " B " licensees, he is condemning the principles of amateur radio in general.

The " B " licence is not for beginners as Mr. Mason seems to think, it is for amateurs primarily interested in v.h.f./u.h.f. techniques.

Perhaps if Mr. Mason had bothered to listen properly on these bands, and heard the standard of operation by G8's (which is in no way inferior to " $A$ " licensees) he may have thought differently when writing his letter.-G. Jones, G8AXE (Liverpool).

## And again!

I am afraid I cannot agree with Mr. Mason's views towards class B licensees in the July issue of Practical Wireless. There seems to be a general feeling of prejudice against class $\mathbf{B}$ licence holders. Perhaps the term " $B$ " conjures up an illusion of their being secondrate operators. This could not be further from the truth because the vast majority of them are experimenters in contrast to a large number of class A licensees who think of their rigs as a telephone substitute. I hope the time has arrived when those holding $\mathbf{B}$ licenses will be permitted to spread their wings and cease to be regarded as outsiders and inferiors.
Mr. Mason strongly implies that a great deal of harm would be done if class B licensees were allowed to use the bands already occupied by those holding class A licenses. He gives no reason why this should be so but I think it might be quite a good idea if A and B were to merge. Bearing in mind the fact that radio amateurs are supposed to be experimenters and not telegraph operators I feel that the morse test should be abolished. A great deal of good would result on the present overcrowded bands if the energy formerly absorbed by the morse test were diverted towards a raising of the standard of the theory examination. With such a provision co-ordination between A and B licences would result.-T. Wright (N. Ireland).

## . . . and even more

Since the ex-P.M.G. Mr. Edward Short said that he was going to introduce a Beginner's Licence all the comments about it have been against such a licence.
As a keen S.W.L. I hope to get my ticket some day but at the moment I am tied up with "O" level studies, making it impossible to study for the R.A.E. and morse test.
It would therefore be very useful to hold a Beginner's Licence and it would also help me when I take my R.A.E. in a few years' time.Jonathan Waters A5438 (London, S.12).

## Solid state

In a valved piece of equipment the actual "work" is done on an electron stream moving in a nearvacuum. The contents of a valve (other than the electrodes) are in the "gaseous-phase or state": electrons and residual air molecules.

When equipment is transistorised, the "work" is done inside a crystal of silicon or germanium. By definition, a crystal is solid; therefore the transistors are in the solid-phase or state. A physicist would say "solid-phase", an electronics bod would say "solid-state".

Incidentally, an electrolytic capacitor works in the liquid-phase! -L. Collier (Billingham, Teesside).

## I like it!

I reply to Mr. Meachim's letter in your September issue of the P.W. I must say I feel sorry for him if he has to listen to experiment Hams Giggling and Squeaking and long drawn out conversations over Oriental Dishes and the removal of warts.

As a short wave listener myself, when I switch on I am prepared to sit back and listen to what they talk about. I find some of their topics very interesting and hope one day to join in the fun, so keep up the good work boys and whether the topic be the removal of warts, how to make a hot pot, or the birth of a baby, Big Brother will be listening. Of course, if one dislikes these topics, the answer is simpleswitch off. Hi!-J. Bennett (Nuneaton).

# trancisitorised SIGNAL GENERATOR JAMES HOSSACK 

FOR the serious experimenter who is also blessed with a well-equipped laboratory and unlimited funds, a high-quality signal generator is a necessity. Such an instrument may incorporate ultra-wide frequency coverage, switched modulation, calibrated signal output, provision for scope synchronisation, etc., and, according to its complexity, entail a corresponding depletion of the available monetary budget.

To those enthusiasts of more limited resources who still feel the occasional need for a test instrument which is capable of injecting a signal, of reasonably reliable frequency, into the household transistor, veteran push-button all-wave radiogram, car radio, or even the domestic TV when required to do so, the present instrument fulfils a useful purpose at modest cost.
In the re-alignment of i.f. stages of old-type valve receivers, it matters little whether the actual i.f. is 440 or $470 \mathrm{kc} / \mathrm{s}$, so long as the various circuits are tuned to the selected frequency in such a way as to give maximum gain consistent with reasonable passband and freedom from "peaking" or instability and "laboratory" accuracy is not required. Similarly the frequency of $10.7 \mathrm{Mc} / \mathrm{s}$ nominally used for i.f. amplification with an f.m. receiver implies a central value located at any point between say 10.3 and $11 \mathrm{Mc} / \mathrm{s}$ (too great a deviation will, of course, result in serious interference from short-wave commercials in the 25 or 31 metre bands) together with a reasonably fiat response, as indicated by a metered output, for some $200 \mathrm{kc} / \mathrm{s}$ either side of this point. Again, it may be desired to line-up a home constructed


Fig. 1: Circuit of a se/f-oscillating mixer, as used in many f.m. receivers.

receiver from scratch. For any of these purposes a relatively simple instrument, on the lines of that described here, will serve the purpose admirably.

## Circuit Description

Originally, the author decided to experiment with a transistor oscillator circuit on the lines of the valve generator described in the August, 1968 issue of Practical Wireless. In this case, signals of the correct phase to maintain oscillation were fed back over a double-triode valve, the second grid being grounded directly to chassis.

It soon became apparent that a single transistor can fulfil the same function when its base is effectively grounded, and feedback initiated between collector and emitter by simple capacitive coupling. This is, in effect, similar to the circuit used in the self-oscillating mixer stage of many f.m. receivers, and is illustrated in Fig. 1.

Unfortunately, to oscillate satisfactorily at low frequencies, a fairly large value for C 3 is required, and this, in turn, imposes such heavy damping on the transistor that oscillation fails to be maintained over the full range available with C2. The position is improved considerably if a high L/C ratio is used (for example, by inserting a ferrite rod into the tuning coil) but, as quite a long rod is required at the lowest frequency range envisaged, and also since portability was considered very desirable, a different approach was adopted. This consists of the addition of a small auxiliary feedback coil for the lowerfrequency inductances, and, since feedback between collector and emitter is substantially in-phase as previously indicated (witness the use of direct capacitive coupling in Fig. 1), the normal tapped coil is unsuitable. This point will be referred to later in connection with coil construction.

Coming now to the audio oscillator used for modulation, and also to provide an output for a.f. test purposes; a more conventional arrangement of oscillator circuit was found to be suitable in this case.


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Generally, any tapped audio coil will provide a continuous a.f. signal if connected between transistor base and earth, with tap connected to the emitter and collector grounded. The author has found that intervalve transformers, tapped chokes, and even transistor output transformers can be induced to oscillate freely, provided careful attention is paid to base biasing arrangements. In the present arrangement a miniature driver transformer was used, the primary and secondary windings being joined at one end so that the feedback direction is correct to maintain oscillation. The correct connections are best determined by trial and error before construction is commenced, otherwise about half the completed assembly may have to be dismantled in order to reverse the connections when it is finally and frustratingly discovered, after switching on the generator for the first time, that a.f. modulation is non-existent.

Figure 2 shows the complete circuit, and a few points are worthy of special mention. First of all, due to the position of the main variable capacitor VCl , which forms part of the collector load of Tr 2 , it was decided to use a negative earth throughout, so that, diagrammatically, both transistors appear inverted. Secondly, the only switch required (apart from the on/off switch which is combined with the output potentiometer) is the 2 -pole 6 -way range switch SW1, which is designed also to function as an r.f.-a.f. switch, providing a.f. only at the output point, when in the correct position. This restricts the number of r.f. ranges to five, and also limits the upper-frequency excursion of the generator, due to the fact that the coils have a common connection to the collector at one end, the earth end only being switched. If the collector end were switched, stray capacitances would be appreciably reduced, but a.f. would not be obtainable without additional circuitry. On the other hand, the highest frequency available could probably be doubled, an additional coil range could be added in the sixth switch position, and the additional a.f. switch necessary could be designed to include an unmodulated radio frequency output also, if required.
Whilst in the present design these advantages have been sacrificed in the interests of convenience and portability-bearing in mind, for example, that an unmodulated output is unlikely to be required for simple indication tests, as opposed to accurate laboratory ones-it is not at all difficult to incorporate all the above modifications into the basic design if this is tackled at the outset, and in fact provision is made on the board layout for a sixth coil.
One further point worth commenting on at this stage is the provision of an additional output socket. This connects directly to the fixed vanes of the tuning capacitor, and enables capacity tests to be

## $\star$ components list

## Resistors:

| R1 $\quad 47 \mathrm{k} \Omega$ | R4 | $1.5 \mathrm{k} \Omega$ |  |
| :--- | :--- | :--- | :--- |
| R2 $\quad 10 \mathrm{k} \Omega$ | R5 | $200 \Omega$ |  |
| R3 $\quad 47 \mathrm{k} \Omega$ |  |  |  |
| All | $\frac{1}{4}$-watt $10 \%$ |  |  |
| VR1 | $500 \mathrm{k} \Omega$ |  |  |

Capacitors:

| C1 | $0.01 \mu \mathrm{~F}$ | C 5 | 120 pF |
| :--- | :--- | :--- | :--- |
| C 2 | $0.01 \mu \mathrm{~F}$ | C 6 | 10 pF |
| C 3 | 0.0018 F | C 7 | $0.001 \mu \mathrm{~F}$ |
| C 4 | 33 pF |  |  |
| VC1 | 400 pF variable air-spaced |  |  |

Transistors:
Tr1 OC71
Tr2 AF115

## Miscellaneous:

T1 midget driver transformer (Henry's Radio); Sw1a/b 2-pole 6-way rotary (Henry's Radio); Sw2 d.p. on/off switch; metal cabinet; 2 coaxial sockets; Veroboard; Perspex; $4 \times \frac{1}{4}$ in. ferrite cores; $2 \times \frac{1}{2} \mathrm{in}$. pointer knobs; $1 \times 1 \frac{1}{4} \mathrm{in}$. knurled knob with cursor; PP3 battery.


Fig. 2: Complete circuit of the transistorised signal generator

## Construction

The container used in the prototype was actually an iron-clad junction box measuring $4 \times 4 \times 1 \frac{1}{2} \mathrm{in}$. (inside dimensions). Any similar type of metal box is suitable, for example a die-cast aluminium chassis of the above dimensions would serve admirably, although a metal rear cover would be needed to avoid unwanted radiation, and also to hold the battery in position.

Commence by drilling the front of the box to take the variable capacitor VCl (centre), together with the range switch, output control, and two output sockets, each located at one of the four corners of the front panel. The "works" proper are constructed in two sections. Firstly, the modulation portion (OC71, transformer T1, and associated components) are built on to a small piece of Veroboard $\frac{3}{4} \times 1 \frac{1}{4} \mathrm{in}$. Since room is available here, it is worth including the AF115 base lead and base biasing components as well, cutting the former lead to a length of about ${ }^{\frac{1}{2}-\frac{3}{3}} \mathrm{in}$. before soldering in position. The Veroboard layout is shown in Fig. 3. Now, place this unit on one side and obtain a rectangle of $\frac{1}{8}$ in. thick Perspex, $2 \frac{1}{4} \times 1 \frac{1}{2}$ in. to hold the coils and other r.f. components, as well as the completed Veroboard unit, which will be mounted in position as described later.

A few words about the use of Perspex as a working material is perhaps relevant at this point. While Veroboard could, of course, have been used throughout, slightly greater difficulty would have been encountered, both with the coil mounting, and in fixing the completed generator in position inside the cabinet, bearing in mind the very small clearances available for inserting fingers, pliers, or tweezers, in order to fit nuts in position, etc. (Constructors who can boast fingertip diameters of less than $\frac{1}{2}$ in. are clearly at a considerable advantage when making up units of this type!) One further advantage of Perspex is its ease of working, particularly as regards threadtapping. All that is required here is to drill a hole slightly smaller than the diameter of the bolt, insert the latter into this hole (most bolts are slightly endtapered, so that the first thread will jam in tightly), and heat the bolt with a soldering-iron applied to its centre. After a few seconds, the Perspex will begin to soften slightly, and if the bolt is now screwed carefully clockwise, then slightly anti-clockwise, in the manner of a tap, repeating until the hole is as


Fig. 3: Veroboard wiring and layout (copper strip side).
deep as required, a perfect thread will be formed when the Perspex has cooled. This trick is used (a) to mount the coils, by tapping threads for 4 in . ferrite cores several threads deep only, using a brass screw of the same thread pitch, and (b) to position fixing screws for the Perspex itself and for the other components including the Veroboard unit, by tapping either all the way through, or half-way through only, depending upon whether the connection or fixing bolt has to be insulated from chassis. The actual template used in these operations is illustrated in Fig. 4.


Fig. 4: Coil template made from Perspex (see text). Drilling dimensions are given below.

A Hole drilled and partly tapped from top for $\frac{1}{4} i n$. cores; B hole drilled and partly tapped from top for 6BA headless bolt; C clearance hole for 6BA bolt to hold Perspex and also take soldering tag for earth connections; D hole drilled and fully tapped for 6BA bolt to hold Perspex; E hole drilled and partly tapped from top for $8 B A$ headless bolts to mount Veroboard on the Perspex; F wire connecting posts (see text).

When drilling and tapping are completed, the ferrite cores are screwed into position and the Veroboard unit can then be mounted, using the 8BA headless bolts-note that the latter must be insulated from earth in the way already described.

One point concerning the Perspex board is the provision of wire "connecting posts", prepared by pressing $\frac{3}{3} \mathrm{in}$. lengths of $18 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. heated copper wire about half-way into the Perspex in the positions shown in Fig. 4. These will remain solidly in position when cold, and are useful for anchoring the fine wire ends of the coils, and for the transistor connections. Ensure that the correct coil ends are soldered to the appropriate posts. However, take care not to overheat when soldering, otherwise they may soften, and come away from the Perspex. Alternatively, use a heat shunt in the same manner as when soldering a transistor connection.

## Winding the Coils

We now come to the rather exacting operation of coil winding, one requiring a fair amount of time, patience, and manual dexterity (see Table 1).

The coil formers are very simply made from cartridge paper or thin card wrapped once around a ferrite core and Sellotaped to form a fairly loosefitting cylinder over the core, approximately $\frac{1}{2} \mathrm{in}$. long. A further thin band of Sellotape at either end forms "end-cheeks", and leaves a central recess about $\frac{3}{8} \mathrm{in}$. wide, just over $\frac{1}{4} \mathrm{in}$. diameter and about

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| 1LD5 | 5/- | fG6a 2/6 | $12 \mathrm{K8GT} 7 / 8$ | 956 2t- | EA50 1/8 | EL41 9/3 | PC900 8/8 | UB41 10/6 | -64 5/6 | BFY50 8/- | OAZ206 9/- |
| 1LN5 | 4/6 | 6H6GT 1/6 | 12Q7GT 3/6 | 1821 10/6 | EA76 13/- | EL42 9f- | PCC84 6/- | UBC4] $7 / 6$ | $\times 65$ | BFY51 4/6 | OAZ20710/6 |
| INSGT | $7 / 8$ | 6J5G 3/9 | 128A7GT6/9 | 5763 10/- | EABC80 6/- | EL81 8/- | PCO85 818 | UBC81 6/6 | $\times 66$ | BFY52 5/- | OAZ210 7/- |
| 1P1 | $71-$ | 6J5GT $4 / 6$ | 125 C 7 4/- | 6060 6/- | EAC91 3/3 | EL33 8/8 | PCC88 11/- | UBF80 $5 / 9$ | X $76 \mathrm{M} 7 / 8$ | BF154 5/- | OAZ213 7\% |
| 1 Pl 10 | $4 / 9$ | $6{ }^{6} 6$ 3/- | 12867 3/- | $719310 / 6$ | EAF42 8/3 | ELS4 4/6 | PCC89 19/6 | UBF89 6/8 | $\times 81 \mathrm{M}$ 29/1 | BF159 8/- | OAZ224 |
| 1P11 | 5/6 | $6{ }^{657 G} 4 / 9$ | 12847 3/- | 7475 4/- | EB34 $7 / 8$ | EL85 7/6 | PCC189 9/3 | TBL21 9]- | X | BF153 4/- | - 16/6 |
| 1R5 | $4 / 8$ | 6J7GT 8/6 | 123578 | A1834 201- | EB41 $4 / 8$ | EL86 8/- | PCF80 6/6 | UC92 5/6 | Y63 26/- | BF167 2/6 | 0C19 25/- |
| 184 | 4/9 | 6K6GT 5/- | $128 \mathrm{~K} 7 \mathrm{~s} /-$ | Ac'044 14/- | E391 2/8 | EL91 8/6 | PCF82 8/3 | CCC84 8/- | Y 63 $5 /-$ <br> Y 65 5 | BF173 2/6 |  |
| 185 | $3 / 8$ | $6 \mathrm{K7G}$ 2/- | 12SQ7GT\%/6 | AC2PES | EBC41 $8 / 6$ | LL95 5- | PCF84 81- | UCC85 616 | Y65 $2 /-$ | BF180 12/- | ${ }_{0}^{0623}$ 14/6 |
| 1T4 | 2/8 | 6 K 7 GT 4/8 | 128 R 7 5/- | 10/6 | EBC81 8/- | EM71 14/- | PCF86 8/6 | LCF80 8/3 | $\begin{array}{ll}263 & 4 / 8 \\ 778\end{array}$ | BF185 8/- | $02414 / 6$ |
| 1U4 | 5/8 | 6 KBG 3/- | 12 Y 4 2/- | AC2PEN/ | EBC90 4/- | EM80 5/9 | PCF801 7/- | 1'CH21 9/- | $\begin{array}{lr}277 \\ 7329 & 11 / 8\end{array}$ | BTX 34/400 | OC25 5 5/- |
| 1U5 | $5 / 3$ | 6K8GT 7/6 | 13 D 1 5/- | D! $19 / 6$ | EBC91 5/- | EM81 7/6 | PCF802 $01-$ | UGH42 ${ }^{\text {O/6 }}$ | Z729 6/- | 401- |  |
| 2 A 7 | 12/6 | 6L6GT 7/9 | 13 D 3 9/- | ACAPEN $4 / 9$ | EBF80 5/9 | EM84 6/- | PCF805 819 | TCH81 6/- | $\begin{array}{ll}2729 \\ \mathrm{Z759} & 88 /-\end{array}$ | BY100 3/6 | OC28 $16 / 8$ |
| 2D13C | $71-$ | $6 \mathrm{LI} 19 / 8$ | $14 \mathrm{H} 7 \quad 9 / 6$ | AC/PEN (5) | EBF83 71- | EM85 11/- | PCF80611/6 | VCL82 7\%- | 2758 88j- | BY101 11/6 | 0c29 16/6 |
| 2D21 | 5/8 | $6 \mathrm{L18} 7 / 6$ | 1487 15/- | 19/6 | EBF89 5/9 | EM87 7/3 | PCF80812/6 | UCL, 83818 |  | BY105 10/6 | $0 \mathrm{OC30}$ 7/- |
| 2 X 2 | $3 /-$ | $6 \mathrm{Ll9}$ 19/- | $18 \quad 12 / 6$ | AC/PEN (7) | EBL22 10/8 | EY51 6/6 | PCLR1 9f- | $\begin{array}{ll}1751 & 9 / 8\end{array}$ | Traneisiors | BY114 $6 / 8$ | 0 O 3510 |
| 344 | $8 / 6$ | $6^{6 L D 20} 6 / 6$ | 19 10/6 | 19/6 | EC52 4/3 | EYR1 7/- | PCLB' ${ }^{\text {8/6 }}$ | CF42 9/- | and diodes | BY126 8/8 | 0C38 718 |
| 3 A5 | $8 / 6$ | 6N7GT 0/6 | 19AQ5 $4 / 9$ | AC/THI101- | EC53 12/6 | EY83 8/3 | PCL® 819 | [F80 6/8 | $2 \mathrm{l} 225510 / 6$ | BY234 4/- | $0 \mathrm{OC38}$ 11/6 |
| 3B7 | $51-$ | $6 \mathrm{Pl} 12 /-$ | $19 \mathrm{Hl} 40 /-$ | AC/TP 19/6 | EC54 $61 /$ | EI'84 9/6 | PCL84 713 | UF85 7/3 | 2 N 404 c | BY236 4/- | $0 \mathrm{OC41}$ 10\%- |
| 3D6 | $3 / 9$ | 6 P 25 12/- | 2001 13/- | AC/VP1121- | ECT0 4/9 | EY86 6j- | PCLES 813 | 17F86 9/- | 2N2297 4/6 | BY238 4/- | OC42 6/8 |
| 3Q4 | 5/3 | 6 626 12/- | $20 \mathrm{D} 480 / 5$ | AC/V P211/- | ECx6 10/3 | EY87 6/- | PCL86 8 8/3 | ${ }^{1} 1 \mathbf{F 8 9}$ 5/8 | $2 \mathrm{~N} 2369 \mathrm{~A} 4 / 8$ | BY Y23 20/- | OC43 12/6 |
| 3Q8GT | $6 / 6$ | 6 P 28 25/- | 20 F 2 14/- | ATP4 $2 / 3$ | EC88 10/3 | EY88 7/6 | PCL88 151- | :141 91- | $2 \mathrm{~N} 312150 \%$ - | BYZ10 5/- | $00^{0} 44 \mathrm{~L}$ |
| 384 | 4/8 | 6Q7G 5/- | $20 \mathrm{Ll} 13 /-$ | AZ1 8/- | Eu92 6/8 | EY91 3/- | PEN45 7\%- | UL46 9/6 | 2N3868 20/- | BYZ11 5/- | O044PM 8/8 |
| 3V4 | 5/6 | 6Q7GT 8/8 | $20 \mathrm{Pl} 17 / 8$ | AZ31 8/8 | ECU31 15/6 | EZ35 5/- | PEN45DD | UL84 6/- | AA120 3/- | BYZ12 5/- | $0{ }^{0} 4518$ |
| 4D1 | $3 / 8$ | 6R7G 5/6 | 20 P 318 f - | AZ41 6/8 | ECC32 4/6 | EZ40 7/8 | 18/6 | UM80 5/- | AA129 3/- | BYZ13 5/- | OC45M 8/- |
| 5R4GY | $8 / 8$ | 68A7ET 7\% | $20 \mathrm{P} 4 \quad 17 / 6$ | B36 $\quad 4 / 0$ | ECC33 20/1 | EZ41 7/3 | PEN46 4/- | UR16 10/6 | AAZ13 3/6 | BYZ15 35/- | OC46 3/- |
| 5U4G | $4 / 8$ | 68A7 7\%- | $20 \mathrm{P5} 17 /-$ | B319 6\%- | ECU34 29/6 | EZ80 319 | P EN 383 9/6 | UC5 7/- | AC107 3/6 | Cal2E 4/- | OC65 $22 / 6$ |
| 6V4G | $7 / 6$ | 68C7GT 6/6 | $25 \mathrm{~A} \mathrm{BG}^{7 / 6}$ | BLF3 10/6 | ECC35 4/9 | EZ81 $4 / 6$ | P'EN384 | CU8 16/6 | ACl13 5/- | Ca64 H 4/- | 0 O 70 2/3 |
| 5Y3GT | $5 / 6$ | $68 \mathrm{G7}$ 6/- | $25 \mathrm{~L} 6 \mathrm{G} 4 / 8$ | CK50n 6/6 | ECC40 9/6 | EZ90 3/6 | 11/6 | [C12 4/6 | AC'14 8/- | GD3 6/6 | 0071 8/- |
| 6Z3 | $7 / 6$ | 68H7 3/- | $25 \mathrm{Y} 361-$ | C14 19/6 | ECC81 4/- | FC4 12/6 | PEN453 D1) | IY1N 10/3 | A('126 2/- | GD4 8/6 | $0 \mathrm{C72}$ 2/- |
| $5 \mathrm{Z4G}$ | 71 - | 68577 5/- | $25 \mathrm{Y} 5 \mathrm{G} \quad 8 / 6$ | CL33 19/6 | ECU82 $4 / 6$ | F W4/5008/6 | - 19/6 | ${ }^{\text {(Y21 }}$ 9/- | $\mathrm{ACl2}^{-}$2/- | GD5 $5 / 6$ | 0073 18/- |
| 6/30L2 | 12/6 | 万SK7GT 4/8 | 25Z46 8J- | CV6 10/6 | ECC83 4/6 | FW4/8008/6 | PENA419/6 | ('Y4) $6 / 6$ | ACliz8 $2 /-$ | GD6 5/6 | 0074 81- |
| 648 G | 5/6 | 68L7GT 4/8 | $25 \mathrm{Z5}$ 7j- | CV63 10/6 | ECC84 516 | (1230 7/- | PEN/DI | 1Y85 $5 / 6$ | AClo4 5/- | GD8 4/- | 0075 2/- |
| $6 \mathrm{AC7}$ | $8 /-$ | 68N7GT 4/8 | 25Z6G 8/6 | CV271 12/6 | ECC85 5/- | G732 $91-$ | 4020 17/6 | (10 8)- | AC155 $6 / 6$ | GD9 4/- | 0076 8/- |
| $6 \mathrm{AG5}$ | 2/6 | 68Q7GT 6/- | $30 \mathrm{C} 1 \quad 6 / 8$ | CW428 19/- | ECC88 7/- | (1733 12/6 | PFL20012/6 | U12/14 7/6 | ACl56 4/- | GD10 4/- | 00778 |
| $6 \mathrm{AG7}$ | $5 / 9$ | 6887 2/- | $30 C 151816$ | CY1 18/4 | Ecc91 3/- | 9:Z84 10/- | PL33 9/- | ${ }_{1} 116$ 15/- | AC157 5i- | GD11 4/- | $0 \mathrm{C78}$ 3/- |
| 6 6AJ5 | 8/6 | 6U4GT 9/6 | $30{ }^{3} 17818 /-$ | CYIC 10/6 | ECO 189 9/6 | (3Z37 14/6 | PL36 9/9 | 117 5/- | AC165 5/- | GD12 4/- | OC78D 81- |
| 6AK5 | $4 / 9$ | 6U5G 5/- | 30 C 18818 | CY31 $7 / 8$ | ECC804 12/6 | H30 51- | PL38 $19 / 9$ | $\begin{array}{ll}\text { C1R/20 } & 6 / 6\end{array}$ | AC166 5/- | GD14 10\% | 0078 8/- |
| 6AK6 | 6/- | 6U7\% 7/- | $30 \mathrm{Fs} 11 / 8$ | 13 1/3 | ECV807 $27 /-$ | HABC808j- | PL81 7/6 | 119 40/- | ACL67 12/- | 4D15 8/- | $0 \mathrm{C81}$ 8/- |
| 6AK8 | 6/- | 6V64 3/6 | 30 FL 1151 - | D15 15/8 | ECF80 6/6 | HL:2 7/8 | PL81A 7/6 | 1722318 | AC168 $7 / 6$ | GDIF 4/- | $0 \mathrm{C81D} 81-$ |
| 6ALE | 2/3 | 6V8GT 6/- | $30 \mathrm{FL} 1215 /-$ | L41 10/6 | ECF'82 6/8 | HL13C 4/- | PL82 5/9 | U25 18/- | AC169 6/6 | 1 1) ET102 4/8 | OC81M 5/- |
| 6AM4 | 16/6 | $6 \times 4$ 3/6 | $30 \mathrm{FLI3}$ 6/- | $1) 63$ 51- | ECP86 9/- | HL23 6j- | PL83 8/- | $\begin{array}{ll}\text { [126 } & 10 / 6\end{array}$ | AC176 11/- | GET103 4/- | OC812 5/6 |
| 6AM5 | 2/6 | $6 \times 5 \mathrm{GT}$ 5j- | 30 FLI4 $12 / 6$ | 1774 | ECF'R0442/- | HL23DD5/- | PL84 6/3 | 1331 6/- | $\mathrm{AC177}^{\mathbf{5} / 6}$ | GET10518/- | perpsir |
| 6AM6 | $3 / 8$ | $6 \times 7 \mathrm{C}$ 12/6 | 30 LI 6/- | DAC32 7- | ECFRO5 18/6 | HL41 3/9 | PL302 11/- | $1 \begin{array}{ll}\text { l33 } & 13 / 6\end{array}$ | ACY17 3/4 | GET113 4/- | $0 \mathrm{OC82} 88$ |
| 6AQ5 | 4/8 | $\begin{array}{lll}\text { 747 } & 12 / 6\end{array}$ | 30 L 151818 | DAF91 3/8 | ECH21 9/6 | H1.410b | PLǒ00 12/3 | V35 16/6 | ACY18 518 | GET1517/- | OC82D $2 / 6$ |
| 6AR6 | 201- | TAN7 6/- | $30 \mathrm{L17} 18 /-$ | DAF96 6/- | ECH3s 5/9 | 19/8 | PL504 13/- | U37 34/11 | ACY19 6/3 | (1ET116 7/8 | $0 \mathrm{CB3}$ 8/- |
| 6AT6 | 4/- | $7 \mathrm{B6} 10 / 9$ | 30P4 12/- | DCC90 8/6 | ECH42 $8 / 6$ | HL42DD8/- | PLK02 15/- | C45 15/6 | ACY20 4/9 | GET119 4/6 | $0 \mathrm{OC84}$ 3/- |
| 6AL6 | 51- | $7 \mathrm{B7}$ 71- | 30 P 4 MR | 1)D4 10/6 | ECH81 5/3 | HN309 $27 / 4$ | PM84 9/3 | 150 | ACY21 $5 / 9$ | (1ET573 8/8 | $\mathrm{OCl}^{\text {OCl }} 3$ |
| 6AV6 | 5/- | 706 61- | 17/6 | DD41 12/6 | ECH83 $7 / 8$ | HVR2 8/8 | PX4 14/- | 1.52 $4 / 9$ | ACY\%2 $3 / 6$ | ¢ET587 8/6 | OC139 12/- |
| 6-84G | 2/6 | $7 \mathrm{DG} 15 /-$ | $30 \mathrm{Pl2}$ 13/- | DDT4 7/6 | ECH84 \%- | HVR2A S/9 | $\begin{array}{llll}\mathrm{P} 31 & 8 / 6\end{array}$ | 1.7648 | $\begin{array}{ll}\text { ACY28 } & 4 / 3\end{array}$ | GET87210/- | OC140 19/- |
| $6 \mathrm{BA6}$ | 4/- | 7H7 5/6 | 30 P19 11/- | DF33 7/9 | ECL80 6J- | [W3 5/6 | PY32 9/6 | $1 * 74$ | AD140 8/- | GET873 4/- | OC169 819 |
| 6BE6 | 4/3 | 7 R 7 12/- | 30 PL 1 15/- | DF72 30/- | EC1.82 6/- | [W $4 / 3505 / 6$ | PY33 9/6 |  | AD149 8/- | GET88210/- | $\begin{array}{lll}\text { OC170 } & 8 / 6\end{array}$ |
| 6BG6G | $20 / 5$ | 787 201- | $30 \mathrm{PL} 13151-$ | DF91 2/8 | ECL83 9\%- | 1W4/5006/- | PY80 5/- | $1: 19112 / 6$ | AF114 4/- | GET887 4/6 | $0 \mathrm{Cl71}$ 3/4 |
| 6BH6 | $71-$ | $7 \mathrm{~V}^{7}$ 5/- | 30 PL14 15/- | DF96 6/- | ECL84 12/- | KT2 5j- | PY81 5/- | 12518 | AF115 3/- | GET889 4/6 | $0 \mathrm{OC172}$ 4/- |
| 6 BJ 6 | 8/9 | $7 \mathrm{Y4}$ 6/6 | 30PL15 15J- | DF97 10/- | ECL85 11/- | КT8 15/- | PY82 5/- | $1{ }^{1} 28181-$ | AF116 3/- | GETP90 4/6 | 0 C 20085 |
| $6 \mathrm{BQ5}$ | 4/6 | 724 | $35 \mathrm{~A} 5151-$ | DH30 15/6 | ECL86 7/9 | KT32 4/8 | PY $\times 3$ 5/6 | $\mathrm{C} 2 \mathrm{R22} 81-$ | AF117 3/4 | GET896 4/6 | OC202 516 |
| 6BQ7A | 7/- | 9BW6 7/- | $35 \mathrm{D} 511 / 8$ | 1 1H63 5/- | ECLRO0 | KT41 19/6 | PY88 6/8 | ${ }^{5} 301111 / 6$ | AF119 3/- | GET897 4/6 | OC203 - 516 |
| $6 \mathrm{BR7}$ | $91-$ | $907 \quad 7 / 6$ | 351669 6/8 | DH76 3/6 | 301- | KT44 5/9 | PY301 12/6 | U329 12/6 | AF124 7/8 | GEX13 3/6 | $0 \mathrm{C} 20410 / 6$ |
| 6BR8 | $81-$ | $10 \mathrm{C1} 12 / 6$ | 35 W 4 4/8 | 1) H 77 4/- | EF'22 12/6 | KT61 12/- | PY800 6/6 | $1 \checkmark 403$ 6/6 | AF125 3/6 | GEX 35 4/6 | OC205 7/6 |
| $6 \mathrm{BS7}$ | 18/6 | 10 C 2 l 10/- | $35 \mathrm{Z3} 10 \mathrm{j}-$ | $1 \mathrm{HH81} 10 /$ - | EF'36 3/- | KT63 4/- | PY801 $6 / 6$ | 14410476 | AF1226 71- | GEX36 101- | $0 \mathrm{C812} 81-$ |
| 8BW7 | 5/6 | $10 \mathrm{Dl} 7 /-$ | $35 \mathrm{Z4GT} 4 / 9$ | DH101 25/- | EF37A 7/- | KT66 16/6 | PZ30 9/6 | ${ }^{\text {L'R01 }} 17 / 6$ | AF127 3/6 | ( ${ }^{\text {EX }} 45 / 17 /$ - | OCP71 87/6 |
| $6 \mathrm{BX6}$ | 4/6 | 10D2 $14 / 7$ | 35Z5ั(:T 5/6 | 1)H107 | EF39 5/- | KT74 $12 / 6$ | QP:21 5/- | $1{ }^{1} 4020818$ | AF'139 11/- | GEX6615/- | ORP12 15/- |
| $6 \mathrm{6BZ6}$ | $81-$ | 10 Fl 15/- | 42 5/- | 16/11 | EF40 8/8 | ${ }^{\text {KT76 }}$ 7/6 | QQVO3/10 | -P4B 11/- | AF178 101- | GT3 5/- | 9x1/6 8/6 |
| 6 C 4 | $2 / 8$ | $10 \mathrm{F9}$ 9/- | 43 101- | LK32 7/- | EF41 9/- | KT88 29/- | 27/6 | ${ }^{17} 130^{\circ} 71-$ | AF179 18/6 | M1 $2 / 10$ | T82 12/6 |
| $6 \mathrm{C5GT}$ | 6/- | $10 \mathrm{~F}^{18} 87 / 8$ | $50 \mathrm{~B} 5 \quad 6 / 8$ | DK40 10/- | EF42 3/6 | KTW61 5/9 | Q575/20 | VP23 2/6 | AF180 9/6 | M3 2/10 | T83 15/- |
| 8 C 6 | 318 | 10LD3 7/6 | 50 Cz 5/8 | 1)K91 4/9 | EF50 2/6 | KTWfit12/6 | 10/6 | $\checkmark \mathrm{P} 41$ 5/- | AF181 14/- | OA5 5/6 | V10/15A |
| 6 C 9 | 12/6 | 101.D11 16\% | 50CDffi41- | I)K92 7/9 | EF54 6j- | KTW63 5/- | Q S $100 / 1 \overline{0}$ | VR75 24f- | AFZ12 5/- | OA9 2/6 | 121- |
| 6CD6G | 19/6 | 10P13 15/6 | 50L6GT 6\%- | DK96 8/6 | EF7S 6/6 | KTZ41 6j- | 9/6 | -R105 5/- | A8Y27 8/6 | OA10 6/8 | MAT100 7/9 |
| fCLI 7 | 9/6 | 10P14 15/6 | $52 \mathrm{Kl}^{+}$14/6 | DL33 6/6 | EF80 4/6 | LN309 8/8 | QV04/7 7/- | VR150 5/- | ASY28 6/8 | OA47 2/- | MAT101 8/6 |
| 6 CH 6 | 61- | 12 A 6 S - | $53 \mathrm{KU} \mathrm{14/6}$ | ${ }^{\text {OLS }} 318$ | EF83 9/6 | $\begin{array}{ll}\text { LPL } & \text { 9/6 }\end{array}$ | R10 15/- | VTfla 7i- | ASY 29 10/- | OA70 3/- | MAT120 719 |
| 6CL6 | $8 / 6$ | 12AC6 7\%- | 72 6/8 | DL7 15/- | EF85 419 | L2319 6/8 | R11 19/6 | VT501 8/- | BA115 $2 / 8$ | OA73 3/- | MAT121 $8 / 6$ |
| 6CW4 | 12/- | 12AD6 6/- | 7750 | Dli5 30/- | EF86 6/- | LZ3 219 9/6 | R12 6/6 | V1'111 6f- | BAll ${ }^{\text {9/- }}$ | OA79 1/8 | ZE12V7 1/8 |

 8.T.C. 1 watt Zener diodes. $2 \cdot 4 \mathrm{v}: 2 \cdot 7 \mathrm{v}: 3 \cdot 0 \mathrm{v}: 3 \cdot \mathrm{fv} ; \mathbf{4} \cdot 3 \mathrm{v}: 33 \mathrm{v}: 16 \mathrm{v}: 18 \mathrm{v}: 30 \mathrm{v}$. All $3 / \mathrm{Bd}$. each.

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[^2]$\frac{1}{t}$ in. deep for the wire itself. It is useful to push the coil former partly over a core while winding, since this enables a better grip to be obtained. Commence with the coil for range 4 (range 5 coil is uncored, and may be self-supporting-see later). This is merely a single winding, consisting of 24 turns of 30 s.w.g. enamelled copper wire wound side by side in the available space, and held in position with Sellotape. Leave approximately $\frac{1}{2}$ in. wire ends for connections.
Now proceed to the double-wound coils for ranges 1, 2 and 3. Two procedures are available here for securing correctly phased windings. The simplest is merely to wind on two separate coils, referring to the data in Table 1, and separating the four ends so that one pair of wires is joined together as in Fig. 5.


Fig 5: Interconnections for the double wound coils.
Note that the windings are in the same direction, and that the start of one winding is joined to the start of the second. The other method is to commence winding a short distance along the wire (approximate distances are indicated in Table 1), leading out the start of this winding to act as a tap. Winding is thus initially bifilar. Wind on, using this double or bifilar mode, the number of turns specified for the auxiliary winding, then lead out the short wire-end (which is, in fact, the true "start" of the wire), Sellotape in position, and continue winding with, of course, one wire, until the total number of turns is equal to that specified for the main winding.

Although it sounds complicated on paper, the latter method is actually very easy to carry out in practice, and enables the rather large amount of very fine wire used, for example, in coil 1 , to be accommodated easily and neatly in the available space. The final operation is to cement the wire firmly in position with coil cement, and when dry, shorten and bare the three connections and test the coil by fitting it over one of the cores (it should slip easily over the latter without, however, undue play) and, if desired, connecting it into a simple transistor circuit to test for oscillation over the specified range. This last step is well worth while. Oscillation should be easily detected on a nearby receiver as heterodyne whistles present over the range in question, or
perhaps as harmonics of double or treble the fundamental frequency.

Now mount all coils in position over the cores, and connect to the wire anchoring leads already described. Wire up the transistor and remaining components, and solder on flying leads approximately $1 \frac{1}{2}$ or 2 in . long for subsequent connection to the range switch.

Two holes should have already been drilled in the side of the cabinet adjacent to the switch to take the fixing screws for the Perspex, and, if tapped threads have been prepared as already described, no difficulty should be experienced in mounting the board in position close to the range switch, as can be seen in the photograph (interior view). Before doing this it is probably wise to check over all connections, and also to fit a small aluminium bracket inside the cabinet to hold the battery, as well as connecting up the output level control and output sockets, since these points may be rather inaccessible once the unit is in position. Finally, solder the flying leads to the switch connections and tuning capacitor, and the generator is ready for testing.


Fig 6: Layout showing the r.f oscillator coils and audio oscillator panel.

## Calibration

Switch first of all to range 6 (extreme clockwise position of the switch), and test that an audio signal is appearing at the output, either with a pair of headphones or by injecting into an amplifier or the a.f. section of a receiver. The output level control should attenuate evenly from a maximum of about 200 mV .

## TABLE 1.

| Coil reference | Range | Approx. frequency coverage | Wire size | Number of turns |  | Approx. length of aux. coil for bifilar winding (see text) | Coil diameter |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Main coil | Auxiliary coil |  |  |
| RC1 | 1 | 200-400kc/s | 40 s.w.g. | 380 | 65 | 6 ft . | To fit |
| RC2 | 2 | 400-950kc/s | 36 s.w.g. | 160 | 35 | 3 ft . | over $\frac{1}{4} \mathrm{in}$. |
| RC3 | 3 | 1000-1800kc/s | 32 s.w.g. | 70 | 20 | 2 ft . | core |
| RC4 | 4 | $7.5-16.0 \mathrm{Mc} / \mathrm{s}$ | 30 s.w.g. | 24 | - | - | (see text) |
| RC5 | 5 | 19-45Mc/s | 20 s.w.g. | . 5 | - | - | $\frac{5}{16} \mathrm{in}$. |

Now switch to range 1 , which is the lowest frequency range. If the coil test previously mentioned has been carried out, the coverage should have been established as approximately correct, and if so, a heterodyne will be heard with the long-wave light programme at $200 \mathrm{kc} / \mathrm{s}$ on an adjacent receiver. This should be close to the low-frequency end of the variable capacitor setting (vanes fully meshed), and, if it is not, carefully slide the range 1 coil out from its core a little, until the $200 \mathrm{kc} / \mathrm{s}$ position is correct. This coil can then be locked in position by spotting with cement, both inside and outside the core.


Internal view of the signal generator.

At this stage, it should be stressed that only a very approximate degree of output attenuation of r.f. is possible with a simple attenuation circuit of the type used here. In fact, unless the metal cabinet is earthed directly to the earth of the circuit under test, and the output taken via a coaxial lead, attenuation will be non-existent since radiation from the cabinet will be picked up more or less directly. This is especially the case for sets with ferrite rod aerials, when strong pickup can occur over several yards. For simple frequency test of the type envisaged (including calibration) this is no disadvantage, but it could be significant, for example, when aligning i.f. stages. Hence, the use of a properly screened and earthed coaxial lead in such cases is recommended.

Now is the time to fit a simple cardboard dial and pointer knob, having first pencilled on two $0-180$ deg. scales on either side of a horizontal line passing through the tuning capacitor spindle. Having determined the $200 \mathrm{kc} / \mathrm{s}$ point accurately, keep the generator tuned to this position, and switch the receiver to medium wave. Find the dial reading corresponding to $600 \mathrm{kc} / \mathrm{s}$ ( 500 M .) This represents the third harmonic of the generator frequency, and sweeping the receiver dial around this area should enable the characteristic tone to be heard, indicating that the receiver is tuned to exactly $600 \mathrm{kc} / \mathrm{s}$. Being a harmonic, it may be necessary to move receiver and generator closer together, or swing the former from side to side, to ensure sufficiently strong pickup. If the tuning point now coincides with the appropriate wavelength on the receiver dial, it may be
assumed that the latter is reasonably accurate, and it ought to be fairly safe to employ it for calibration purposes, at least on the first three ranges. If the fifth harmonic of $200 \mathrm{kc} / \mathrm{s}$, at a reading of $1 \mathrm{Mc} / \mathrm{s}$, is audible, a further check for receiver accuracy can be made. If the receiver calibration is obviously well out, while, in theory, it can still be used by making use of broadcast stations of known frequency, the author's recommendation is to beg, borrow or even buy an accurate receiver-preferably an all-wave one, at the same time making a mental note to re-trim the faulty receiver as and when the signal generator is fully operational.

Returning to the calibration, and assuming that the receiver accuracy is sufficient; move the generator dial slowly in the high-frequency direction, and follow the tone with the receiver dial until $660 \mathrm{kc} / \mathrm{s}$ or the corresponding wavelength in metres-is reached. Mark this spot as $220 \mathrm{kc} / \mathrm{s}$ on the generator dial. Repeat the procedure until range 1 is fully calibrated. Now switch to range 2, and check the approximate coverage. The low-frequency end of this range should be close to $400 \mathrm{kc} / \mathrm{s}$, and using the receiver at the second harmonic point ( $800 \mathrm{kc} / \mathrm{s}=$ 375M.), this reading can be accurately marked on the scale for range 2. Repeat the procedure of range 1 as far up as $550 \mathrm{kc} / \mathrm{s}$; beyond this, the fundamental will prove more useful, as the generator is now within the medium-wave band. When the coverage has been satisfactorily determined and calibration of this range completed, lock the coil on its core, as with range 1 .

Range 3 also covers medium-wave for part of its spread, so a similar drill is carried out, although the extreme high-frequency end may need to be calibrated either by graphical interpolation or by using harmonics falling in the receiver short-wave band.

For range 4, a short-wave receiver is, of course, essential. This range covers the most useful portion of the short-wave band (roughly, 18-40 metres). Because of the different circuitry of the oscillator, however, the output is lower than on the first three ranges, so that direct contact with the receiver aerial (usually a whip aerial in the case of a transistor portable) may be required. A probe made up from an old coaxial plug will prove a handy accessory in this case, and will also be useful later, if it is desired to use the generator as a hand-held signal injector for directly probing around inside faulty receivers or amplifiers. Its small size makes it ideal for this application. It is stressed once more however, that the output control will not function satisfactorily in these cases.

One more point about calibration of this rangeit is very easy, with some cheap non-selective shortwave sets, to pick up harmonic or image signals from the generator, mistaking these for the main fundamental. To avoid this, it is essential to carefully examine the readings obtained and, if necessary, enter them on graph paper, as, by this means, any erroneous values will be quickly spotted. As before, lock the coil in position once the calibration has been completed.

No mention has yet been made of the coil for range 5, covering the highest frequencies. This can be quite a tricky range to make operational, particularly since a fairly strong degree of oscillation is needed to provide the harmonics necessary for f.m. and TV testing. On the other hand, the connecting points to this coil are very accessible, and, once the

## DRY JOINT TESTER

The most rellable way of teating for a dry joint is to measure the resistance between the component doing this comprises a large scale (3in.) moving coil doing this comprises a large scale (3in.) moving coil meter, a varisble resistance for adjuating zero The only additional items you will need are a battery, some wire, a pair of test rods. Price 19/6, postage and tnsurance 2/6.

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## BARGAIN OF THE YEAR

## MICRO-SONIC

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 In transit from the East these sets surered silght wiere in them but when teries were left in them but when
this corrosion is cleared away they

Designed to operate transistor sets and smplifiers. 000 mA (ciass B working). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6 PP7, PP9, and others. Kit comprises: maine transformer rectifler, smoothing and load resistor condensers and instructions. Real snip at only 18/6, plus 3/6 postage.


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3. Morse tapper.
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or transmitter. Chrome padio ted-six sections, extends from screw. $7 / 6$ MAINS RELAYS
Type A $210 / 250 \mathrm{~V}$ AC ooil, 2 types, one for single whole chassis mounting, has 3 pairs heavy duty changeover contacts. $8 / 6$ e ch. yype octal valve holder. 2 pairs changeover contacts 12/6 each.

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## ELECTRONICS (CROYDON) LIMITED

(Dept. P.w.) 266 LONDON ROAD, WEST CROYDON CRO 2TH 102/3 TAMWORTH ROAD. CROYDON, SURREY
remainder of the unit is functioning, a few test coils can quickly be made up on the lines indicated by the coil data table (Table 1), and wired across the appropriate points. When a suitable inductance has been found, either solder in position with stout wire, or bolt to the cabinet using a brass (not iron) fixing screw. No core is necessary.

After some trials, the following procedure for calibrating this range was adopted, and proved sufficiently accurate for the purpose. Obtain a v.h.f./f.m. receiver with a dial marked from 88 to $100 \mathrm{Mc} / \mathrm{s}$, and set this to the $88 \mathrm{Mc} / \mathrm{s}$ position. Tune the generator, with a suitable coil in position, over the full range, having first, of course, connected the output to the f.m. aerial terminal of the set, and a number of responses should be heard. Select the topmost position (vanes unmeshed) and note the reading in degrees on the cardboard scale of the generator.
It is probable that the receiver is responding to the second harmonic of $44 \mathrm{Mc} / \mathrm{s}$, so that the latter figure corresponds to the signal generator frequency at the dial reading noted. Mark this point on a piece of graph paper scaled to read degrees against frequency. Now move to the second-top response, which is about $29.3 \mathrm{Mc} / \mathrm{s}$, and the third, which will be $22 \mathrm{Mc} / \mathrm{s}$, and plot these readings. Now retune the receiver to $100 \mathrm{Mc} / \mathrm{s}$ and repeat the above procedure. The tuning points this time should be $33 \cdot 3,25$ and $20 \mathrm{Mc} / \mathrm{s}$, and all six points should lie on a smooth curve or perhaps a straight line. If they do not, it is probable that the coil coverage is incorrect, and either the coil should be replaced, or a new set of readings plotted, assuming different harmonics and testing for graphical linearity as before. Should a coil be wound which is found to generate a strong signal above, say, $60 \mathrm{Mc} / \mathrm{s}$, this could, of course, be used, but the lower end of the range would then in all probability not reach a sufficiently low frequency, leaving a rather wide gap at the important 13 metre band. The combination of frequencies shown for ranges 4 and 5 is, in fact, the most suitable for general receiver alignment.
After calibration has been completed for all 5 ranges, the various readings can be recorded on the instrument dial and checked carefully before being finally inked in. The original degree scale can be left on, if desired.

## Using the Generator

For accurate alignment of the intermediate frequency stages of f.m. and particularly of TV receivers, an instrument of this type is not really adequate, though, on the other hand, very useful indications of circuit continuity can be obtained in many cases, enabling location of faults to be simplified. The r.f. stages of TV receivers can be induced to respond to suitably chosen second and fifth harmonics of the generator frequency in Bands 1 and 3 respectively, while direct injection of a $34 \mathrm{Mc} / \mathrm{s}$ signal will serve to test the i.f. performance.
Audio amplifier faults generally fall into two classes-those in which there is a complete absence of input signal appearing at the speaker, and those in which the signal is present, but is either too weak and/or distorted. In the first case, using the generator as a portable a.f. signal injector should at least localise the fault, and may indicate directly a faulty component in the circuit. In the second case, advantage can be taken of the fact that, at least in the a.f. position, the attenuation control is reasonably linear with output, so that an idea of the gain associated with each stage of the amplifier may be obtained by noting the position of this control which is required to achieve a given volume of output from the speaker.

Finally, test facilities are available at socket 1 for ascertaining the values of small capacitors, up to 400 pF , from which the marking has been removed or obliterated. A separate receiver is required. Tune the generator to the top of range 3 , i.e. fully clockwise, and pick up the a.f. tone on the receiver at about $1 \mathrm{Mc} / \mathrm{s}$-the actual frequency is unimportant. Now place the unknown capacitor in position across socket 1 (it is now effectively in parallel with VC1), and turn the generator tuning anti-clockwise until the tone is again heard on the receiver. The value of the capacitor corresponds to the difference between the first and second readings. If no note is detectable, the capacitor is larger than 400 pF or is faulty. This test scale should be calibrated initially by using capacitors of known value, and, when completed, by adding these markings as a separate range.

## Practical Gift

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# SBOWME 68 

THIS year's radio and television exhibits, held in well-known London hotels during the period August 25 th to 29 th, were for the trade only, like last year's shows. All exhibitors were well pleased with the results of their efforts, some manufacturers selling their next four months' production during the first day or two.

The success of trade-only shows means that a return to a national radio and television exhibition is most unlikely. This seems a great pity in view of the current revival of interest in radio-due to the opening of local stations, the output of "pop" from Radio 1, and the increased availability of stereo programmes.
A national exhibition would also take advantage of the growth of the colour-television market, and show the public that the BBC's colour service is excellent and not to be judged by the poor displays so often seen in shop windows. (A report on the new developments seen in TV receivers will appear in the November issue of the companion magazine "Practical Television", on sale on October 18th.)
One of the first points noticed during our tour of the shows was the increased number of foreign products, mainly Japanese. There were many British items too, but the price and finish of the foreign goods certainly compete with those of home-produced equipment.
More radios and radiograms now have v.h.f. coverage, and a large number now cover one or two short wavebands. Radiograms often have provision for the addition of decoders for stereo radio reception but most makers supply these items as "extras" One wonders how long it will be before decoders are standard in all stereograms.

Cabinet styling this year reflected the current fashion, some units being painted in vivid colours and some being covered in brightly coloured leathercloth (a good example of the bringing-up-to-date of a traditional material). Designers had also obviously been hard at work trying to give an expensive appearance to the cheaper lines. This was achieved by removing much of the chrome glitter to make a more attractive product.

Some makers were pleased with their redesigned control knobs, failing to realise that most of their competitors had had the same ideas-the popular style this year was the spun aluminium knob consisting of a squat cylinder with a bevelled edge.

Tuning scales on radios are now up-to-date, with the "square" names such as "Light Programme" banished. However, the use of terms such as "BBC 1" and "BBC 2 " on a radio is rather puzzling when seen for the first time-we haven't yet heard of a customer returning a radio because of a poor BBC-2 picture, but we are sure it will happen eventually!
A full review of the various exhibitions would
take up too much space and would be useful only to a minority of our readers. We have therefore picked out from the vast number of new items those which we think would be of particular interest.

A range of aircraft and marine d.f. receivers was shown by Perry and Pharo. The Novapal priced at $£ 27$ covers $190 \mathrm{kc} / \mathrm{s}$ to $400 \mathrm{kc} / \mathrm{s} ; 550 \mathrm{kc} / \mathrm{s}$ to $1600 \mathrm{kc} / \mathrm{s}$; and $1500 \mathrm{kc} / \mathrm{s}$ to $4500 \mathrm{kc} / \mathrm{s}$. This receiver comes complete with beacon charts, navigator's handbook, operating manual, beacon lists, radio logs, and carrying case. The Aviator D.F. receiver has a similar specification but also covers $103 \mathrm{Mc} / \mathrm{s}$ to $136 \mathrm{Mc} / \mathrm{s}$ and retails at $£ 53 \mathrm{l} 0 \mathrm{~s}$. 0d. The Action D.F. retails at $£ 53 \mathrm{10}$. 0d. too, and has two v.h.f. wavebands ( $150 \mathrm{Mc} / \mathrm{s}$ to $175 \mathrm{Mc} / \mathrm{s}$ and $30 \mathrm{Mc} / \mathrm{s}$ to $50 \mathrm{Mc} / \mathrm{s}$ ).

Also shown by Perry and Pharo was the Pygmy 2001 which covers f.m., long and medium waves, and short waves from 25 to 190 metres, with four bandspread positions (for the 13, 16, 19 and 25 metre bands).
Sony has a pocket sized a.m./f.m. radio finished in black and silver and weighing only 9 oz . The retail price is $£ 11$. Also shown by Sony were three unusually shaped receivers the TR1819, TR1829 and TR1839, respectively cylindrical, cubic, and cigar-box shaped, as may be seen from the illustrations. They retail at $£ 67 \mathrm{~s}$. $0 \mathrm{~d} ., £ 8 \mathrm{~s}$. 0 d ., and £9 8s. Od.
The HP180 from Sony is a modular stereo system, with an output of 7 W per channel, comprising gram unit, a.m./f.m. stereo tuner, and a pair of bookshelf loudspeaker units. The complete system is $£ 125$.

Sobell announced the SG685 transistorised four waveband stereogram. The cabinet is finished in teak with a black leathercloth interior. The tuner section covers l.w., m.w., s.w., and v.h.f./f.m. and uses silicon planar transistors.

The G.E.C. 836 is a new a.m./f.m. portable covering l.w., m.w., m.w. bandspread, v.h.f./f.m., two s.w. bands and two s.w. bandspread. This model has a time-zoned map of the world on the back panel and a tuning meter which is also used to check the battery. A turntable is built into the base of the set.

A clock radio from Ferguson to retail at f17 10s. Od. combines a seven-transistor radio and a battery-operated clock which can be set to switch on the radio at any desired time. The model-number is 3163.

Portable model 6151 from Ultra covers I.w., m.w. and v.h.f./f.m and is finished in dark grey and chrome. The price is $£ 155 \mathrm{~s}$. Od.
A new item from Discatron, the TR3000 Tri Combo, is a combined radiogram and cassette tape player. The record-player plays any 45 r.p.m. record and uses the now well-known Discatron linear tracking principle. The unit features a 7 in . x 4 in . loud-


Sobell SG685


Eagle RM9
"Rhythm Master"


Elsworthy Electronics "Interceptor'


KB KP038 record player with stand (extra) and KA039 matching amplifier


Dynatron HFC8


Murphy B845
speaker and full coverage of m.w. and l.w. The recommended retail price is $£ 283 \mathrm{~s} .5 \mathrm{~d}$.
Aiwa were showing the TP. 1018 tape cassette stereo auto-changer. This unit is mains operated and will play six cassettes one after the other automatically, giving up to $4 \frac{1}{2}$ hours playing-time (one side of each cassette). The units weigh $20 \frac{1}{2} \mathrm{lb}$. and will be a vailable next year
An unusual item from Aiwa was a radio tuner unit in the form of a tape cassette which thus converts a cassette tape player into a radio. This unit retails at less than $£ 5$ and covers $m . w$

A four-track stereo tape recorder was also shown by Aiwa (Model TP.1012) and can be operated from a.c., d.c., or a car battery.

Fidelity had a new tape recorder, the Braemar, which is available in a four-track version for 32 gns., and a two-track version for 29 gns . This is a $3 \frac{3}{3} \mathrm{in} . / \mathrm{sec}$. machine with $5 \frac{3}{4}$ in: spools. The Fidelity RAD 14 is a new portable radio covering l.w., m.w., and v.h.f./f.m. It weighs $4 \frac{1}{2} \mathrm{lb}$., costs 17 gns ., and is finished in green leathercloth.

The Interceptor from Elsworthy Electronics covers m.w., v.h.f./f.m. and short waves ( $6 \mathrm{Mc} / \mathrm{s}$ to $18 \mathrm{Mc} / \mathrm{s}$ ). However, it also covers $142 \mathrm{Mc} / \mathrm{s}$ to $150 \mathrm{Mc} / \mathrm{s}$, which includes the 2 metre amateur band. The set weighs 7 lb . and works from batteries or mains. It has 18 transistors, slide-rule tuning, and automatic circuitry to switch to battery operation if the mains supply fails. The retail price of this receiver is 39 gn .
Bosch Bluespot introduced four new radiograms: the Stockholm at 115 gns .; the Santiago also at 115 gns.; the Arkansas Mk. IV at 139 gns.; and the Barcelona Mk. IV at 155 gns . All include automatic stereo decoders and the Barcelona covers v.h.f./f.m. and three short wavebands one of which is the marine band. The other three radiograms cover l.w., m.w., and v.h.f./f.m.
Teleton Electro were showing a solid-state stereo a.m./f.m. receiver, model 7AT-1. This has an f.e.t. in the f.m. front-end, and covers $88 \mathrm{Mc} / \mathrm{s}$ to $108 \mathrm{Mc} / \mathrm{s}$ on $\mathrm{f} . \mathrm{m}$., and $565 \mathrm{kc} / \mathrm{s}$ to $1605 \mathrm{kc} / \mathrm{s}$ on a.m. The tone controls are comprehensive and high-pass and low-pass filters are included. The suggested retail price is $£ 1188 \mathrm{~s}$. 7 d .
The KTR-1381 from Koyo is a new receiver covering m.w., l.w., s.w. ( $4 \mathrm{Mc} / \mathrm{s}$ to $12 \mathrm{Mc} / \mathrm{s}$ ), and v.h.f./f.m. There are 13 transistors and the price is 25 gns . The Koyo TKR-1651L has 16 transistors and features a built-in mains adaptor. In addition to v.h.f./f.m.. m.w., and l.w., the coverage includes two short wavebands, $3 \mathrm{Mc} / \mathrm{s}$ to $9 \mathrm{Mc} / \mathrm{s}$, and $9 \mathrm{Mc} / \mathrm{s}$ to $22 \mathrm{Mc} / \mathrm{s}$. The retail price of this model is $37 \frac{1}{2} \mathrm{gns}$.

Mordaunt-Short were showing seven loudspeaker systems ranging in price from 33 to 65 gns . The MSIO0 at 33 gns , is a bookshelf type rated at 25 W music power. This unit features three loudspeakers, including the Decca Kelly 8in. flat piston. and a specially developed E.M.I. mid-range unit. The MS700 is a floor-standing system which has a bass unit and a Decca-Kelly ribbon unit. An acoustic lens is also fitted to give wider sound dispersion.

All of the Mordaunt-Short systems are designed on the infinite-baffle principle and the MS100, 200, 300 and 400 use ferrite-cored cross-over inductors in conjunction with paper capacitors.

The Dynatron HFC8 tuner/amplifier record-player covers three wavebands and includes a stereo deco-
der. There is a choice of a Garrard auto-changer or single-player at retail prices of $£ 92 \mathrm{os}$. Od and $£ 94$ 10 s .0 d . respectively.
Pye's International 3042 has 11 transistors and covers l.w., m.w. and seven s.w. bands, five of which have bandspreading. This receiver operates from a.c. or d.c. mains and is priced at $£ 410$ s. 0 d .

Also on view was the Pye "Picadilly" 6000 which covers l.w., m.w., v.h.f./f.m., and s.w. from $1.6 \mathrm{Mc} / \mathrm{s}$ to $27 \mathrm{Mc} / \mathrm{s}$ in four bands. Electronic bandspreading is featured and the receiver is powered from six 1.5 V cells.

Highlight of the Sinclair show was the new IC10 integrated circuit 10 W amplifier. This is priced at 59 s . 6 d . including instruction manual containing a number of useful circuits and ideas. The size of the unit is only lin. x 0.4 in . x 0.2 in ., but the output is 5 W r.m.s., with a frequency response from $5 \mathrm{c} / \mathrm{s}$ to $100 \mathrm{kc} / \mathrm{s}$. To convert the IC10 into an audio amplifier, only tone and volume controls are needed. The power supply can be 8 V to 18 V .

A new five waveband radio from Murphy, the B845, covers m.w., l.w., v.h.f.ff.m. and s.w. from about $5 \cdot 5 \mathrm{Mc} / \mathrm{s}$ to $22 \mathrm{Mc} / \mathrm{s}$. A logging scale is provided and there are separate bass and treble controls. The retail price of this model is 38 gns .
The RR380 from Philips combines a l.w./m.w. portable radio and portable cassette tape recorder. Automatic recording-level control gives optimum results from microphone, pick-up, or radio. The price of the complete unit is 36 gns .

Alba showed their 666 a.m./f.m. portable which uses nine transistors and five diodes. The retail price of this model is $£ 1710$ s. 0 d.

Telefunken exhibited a new version of their stereo tape recorder 204. The new three-speed 204TS has VU meters, a transparent cover, and the carrying handle is now at the top. The M250 hi-fi is a twotrack stereo recorder operating at $7 \frac{1}{2} \mathrm{in} . / \mathrm{sec}$. There are facilities for echo and the price is 150 gns . Matching audio units will be available later.

Denham and Morley showed two new table mains radios. The "Maestro" with seven valves covers four wavebands-l.w., m.w., v.h.f./f.m., and s.w. ( 15 to 50 metres). The a.m. and f.m. tuning are separate and sockets are fitted for tape recorder, pickup, external aerials, and extension loudspeakers. The "Melody" covers I.w./m.w. and v.h.f./f.m.
Radionette showed the Kurer 1001, a twelve-transistor a.m./f.m. portable with press-button controls. Two telescopic aerials and separate bass and treble controls are features and the receiver retails at 39 gns .

The "Bolero Stereo 40 " from B.M.B. is a stereo record-playing system having an output of 5 W per channel, A Garrard 3000 LM auto-change deck is fitted and has a Sonotone 9TAHC stereo cartridge. The system will retail at $69 \frac{1}{2}$ gns.

The KP038 from ITT K-B is a portable record player finished in natural teak with black and silver grille. A matching amplifier (KA039) is available for stereo, and a pedestal stand is an optional extra. The KP038 retails at 34 gns. and the KA039 at 15 gns .

Eagle products showed a very interesting item, the RM9 "Rhythm Master" which electronically generates the sounds of percussion instruments ranging from the bass drum to the cymbal, and produces the sounds in nine rhythms including waltz and bossanova. The unit operates from 240 V a.c. mains and the output is suitable for feeding to an amplifier.


## UNIJUNCTION TRANSISTOR CIRCUITS c.r.bradley

TYHE unijunction transistor differs from the ordinary $\mathrm{p}-\mathrm{n}-\mathrm{p}$ and $\mathrm{n}-\mathrm{p}-\mathrm{n}$ transistors in having only a single p-n junction. It is very similar to the junction diode with the difference that the $\mathrm{p}-\mathrm{n}$ junction is made near one end of a bar of n-type material with leads at both ends called the base 1 and base 2 connections. The circuit symbol and an equivalent circuit for the unijunction transistor are shown in Fig. 1. The diode in the equivalent circuit represents the p-n junction and its anode lead is called the emitter connection.


Fig. 1: (a) Unijunction transistor circuit symbol. (b) Equivalent circuit of a unijunction transistor.

If base 1 is earthed and a positive voltage Vbb applied to base 2 a small current flows due to the resistance of the n-type bar ( 5 to $10 \mathrm{k} \Omega$ ). The voltage at point A will be a fixed fraction of Vbb viz: $[R b 1 /(\mathrm{Rbl}+\mathrm{Rb} 2) \mathrm{Vbb}]$. When Ve is less than this voltage the diode is reverse biased and practically no emitter current flows. But if Ve exceeds this emitter current will flow. This injects holes into the bar which act as current carriers and the resistance of the bar drops to a very low value.

The transistor may be used as an on/off switch, the on condition being:

$$
\text { Ve greater than } \frac{V b b R b 1}{(R b 1+R b 2)}
$$

The main characteristics of the unijunction transistor as a switch are: (1) Fast switching time with low emitter current. (2) The ratio $\mathrm{Rb} 1: \mathrm{Rb} 2$ is fixed by the construction of the transistor and the on condition is virtually independent of temperature. (3) High peak current capacity. (4) Unlike a thyristor the unijunction may be switched off by removing the triggering (emitter) voltage.

## Relaxation Oscillator

The oscillator circuit shown in Fig. 2 has a variety of uses and operates as follows. At the start assume that the emitter is at earth potential and the unijunction is off. The capacitor C now slowly charges through resistor $R$. When Ve reaches the firing voltage the
emitter suddenly conducts and the capacitor discharges through the unijunction. Current also flows from base 2 to base 1 and is limited to a safe value by R1. The emitter current falls very quickly as the capacitor discharges and the unijunction switches off again. The cycle of operations is then repeated and the circuit continues to oscillate at a frequency fixed by the values of the RC combination. A sawtooth waveform is available at the emitter and short negative-going pulses may be taken from base 2 .


Fig. 2: Unijunction relaxation oscillator. The frequency of oscillation is determined by the values of $R$ and $C$.

The frequency of oscillation may be varied by adjusting R and C . For high frequency oscillation use a low value for $R$ and a small one for $C$; conversely for low frequencies use a large $R$ and a large $C$ value. A very wide range of frequencies is possible but R and C must lie within the following limits:
(1) If the value of $R$ is too high it cannot supply enough current to switch the unijunction on. It also has to supply any leakage current drawn by C which, if possible, should be a low-leakage (i.e. non-electrolytic) type. The maximum value of R with which a 2 N 2646 unijunction would oscillate reliably was found to be about $1 \cdot 5 \mathrm{M} \Omega(\mathrm{Vbat}=30 \mathrm{~V})$.
(2) If the value of $R$ is too low it will supply the triggering current continuously and the unijunction will not switch off. If the base-base current is not limited sufficiently by R1 the unijunction will be destroyed. Under the same conditions as above the minimum safe value of R with a 2 N 2646 was found to be about $1 \mathrm{k} \Omega$.
(3) If C is too large the peak discharge current may be damaging to the unijunction. For C values above a couple of $\mu \mathrm{F}$ a resistor of at least $1 \Omega$ per $\mu \mathrm{F}$ should be put in with series C.
These limitations are not very severe and (2) is the main one that has to be borne in mind-otherwise one will waste a lot of unijunctions!

## Code Practice Oscillator

The audio oscillator shown in Fig. 3 is designed for morse code practice and is an ideal exercise for the


Fig. 3: Unijunction code practice oscillator. VR1 is the tone control. S1 provides speaker muting.

## components list

```
R1 680\Omega }\frac{1}{4}\textrm{W},10
R2 3.3k\Omega \frac{1}{4}W,10%
VR1 20k\Omega carbon pot
C1 0.47 or 0.5 F paper
S1 s.p.s.t. slide or toggle switch
SKT Closed circuit jack
LS 3-15\Omega speaker, any size
Tr1 2N2646 or equivalent
```

Three PP3 type batteries, terminal strip, battery clips, wire etc
beginner, It is so simple that it can hardly fail to work first time! There is no need for an on/off switch as the circuit draws no current when the morse key is not depressed. If a pair of test leads are connected across the key as shown the device will double as an audible continuity tester. The layout shown uses a break-off terminal strip and the whole instrument could be wired without a soldering iron.

## Unijunction Metronome

This simple application of the unijunction oscillator is useful for any kind of music practice. It gives loud rhythmic beats similar to a mechanical metronome. The circuit is shown in Fig. 4 and like all these circuits will work with unijunctions other than the 2N2646 specified (e.g. UT46, 4JD5E29, TIS43, BEN3000) although the timing components may have to be adjusted. If doing this double the value of R1 and put a warning milliammeter in series with the battery for protection against condition (2) above while experimenting.
The values of $\mathrm{C} 1, \mathrm{R} 2$ and VR1 were selected to give


Fig. 4: Unijunction metronome. VR1 gives a range of 90-220 beats per minute. An output suitable for feeding into an amplifier is available from the headphone socket. A suggested lavout is given overpage.


Suggested layout for the unijunction code practice oscillator.
a range of 90 to 220 beats per minute, suitable for most musical purposes. VR1 should be an ordinary logarithmic carbon potentiometer wired to give decreasing frequency with clockwise rotation. This is to avoid a compressed scale at the high-frequency end. The author used a pair of $2 \mu \mathrm{~F}$ electrolytics for Cl as these were more readily available than a $4 \mu \mathrm{~F}$ unit, but a $4 \mu \mathrm{~F}$ paper or electrolytic type would be equally suitable.

The metronome will give loud beats that can be heard above most musical instruments. If a 12 in . speaker is used there should be ample volume, or several speakers can be wired in parallel as shown. Headphones can be used for loud pop music practice although in this case the metronome beats may be injected into a guitar amplifier (see Fig. 4). Volume may also be increased by increasing the supply voltage although this experiment is not advised unless one is prepared to destroy a unijunction finding out its maximum ratings!

As the batteries wear out the volume drops but the beat frequency remains virtually constant. Layout is not critical and the circuit may be wired on the same lines as the code practice oscillator. Fit a large pointer knob to VR1 and draw a calibrated scale $90-220$ using a stopwatch. The range actually obtained may differ from this slightly due to component tolerances and slight adjustment of R2 value may be necessary. Increasing it to $82 \mathrm{k} \Omega$ will give a slower range of beats.

## components list

R1 $680 \Omega \frac{1}{4} \mathrm{~W}, 10 \%$
R2 $68 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W}, 10 \%$
VR1 $100 \mathrm{k} \Omega$ carbon pot
C1 $4 \mu \mathrm{~F} 30 \mathrm{~V}$ or above electrolytic
S1 s.p.s.t. slide or toggle switch
SKT Open circuit jack
LS $3-15 \Omega$ speaker, 4 in . or larger
Tr1 2 N 2646 or equivalent
Three PP3 batteries or mains power supply, terminal strip, battery clips, wire, etc.


Suggested layout for the unijunction metronome.
The metronome may be run from three PP3 batteries (27V) or from the mains power supply circuit shown in Fig. 5. Battery life should be 2-3 months with normal use.


Fig. 5: Simple unregulated power supply suitable for use with all the circuits given in this article. Extensive smoothing is not required.

## components list

| R1 | $22-50 \Omega \frac{1}{W} \mathrm{~W}$ |
| :--- | :--- |
| C1 | $500 \mu \mathrm{~F} 50 \mathrm{~V}$ electrolytic |
| D1 | Silicon rectifier, 50 V p.i.v. |
| S1 | Mains d.p....t. on/off switch |
| T1 | Mains transformer, 25 V a.c. secondary |

## Monophonic Electric Organs

These organs are often seen in these pages and need not be described at length here, the idea being to run an audio oscillator at musical frequencies selected by a keyboard (or see Practical Wireless July-September


Fig. 6: Unijunction organ oscillator. VR1-VR37 are tuning resistors for each note. The circuit will drive a speaker directly at medium volume.

## * components list

R1 $680 \Omega \frac{1}{4} \mathrm{~W}, 10 \%$
R2 $1 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W}, 10 \%$
VR1-VR37 $10 \mathrm{k} \Omega$ skeleton presets
C1 0.47 or $0.5 \mu \mathrm{~F}$ paper
C2 $0.1 \mu \mathrm{~F}$ paper
C3 $0.1 \mu \mathrm{~F}$ paper
S1 s.p.s.t. on/off switch
LS $3-15 \Omega$ (if used)
Tr1 2N2646 or equivalent
Three PP3 batteries or mains power supply, keyboard, etc.

1968 for a Portable Keyless Organ). These organs usually use a two-transistor multivibrator oscillator circuit but a unijunction relaxation oscillator could be used with these advantages: (1) A wider range of frequencies is available by varying one resistor. (2) Two different waveforms are available from one oscillator (sawtooth wave from the unijunction emitter-a "coarse" sounding note; and narrow pulses from base 2 -a buzz or whining sound). (3) The oscillator will drive a speaker directly (at medium volume) without amplification. (4) Lower current drain: no current is drawn when no note is played.

A very simple unijunction organ circuit is shown in Fig. 6. This may be used to drive a speaker directly and/ or deliver two different signals to an amplifier. The mechanical details of the keyboard are left to the constructor, as is the layout which is not critical. This is a monophonic organ so that only one note may be played at a time.

TO BE CONTINUED

## COLOUR CONVERGENCE

Find out all about this important aspect of colour television. Separate articles on why convergence correction is necessary, the techniques used and how to carry out convergence adjustments.

> SHOW REPORTS BAND III LOFT AERIAL in the NOVEMBER issue of PRACT/CAL TELEVISION on sale October 18th


No. 2 - TRANSDUCERS

THE transducer is a device which changes one form of energy to another. A pick-up cartridge converts the energy of a moving stylus to minute voltages; microphones receive pressure and velocity variations in the air and produce voltage changes; loudspeakers act in reverse, being powered by current and voltage and moving the air to give sound. Our final example, the tape recorder head can either produce a varying magnetic field when current passes through it, or a current from its windings when a varying magnetic field passes across the gap in its magnet.

The principles are well known, but the problems that arise in audio matters are mostly due to the nonlinearity of the conversions. From an electromechanical point of view, most transducers are "linear" in that the output is proportional to the input but various loss factors enter when we consider frequency conversion and correction is often needed in the electronic circuits to which transducers are coupled to compensate for this "frequencyconsciousness"

Of the four classes of transducer mentioned above, three work normally within the bounds of the classic transducer pattern. Types are: (a) piezo-electric (b) moving conductor (c) variable magnetic field and (d) electrostatic. The exception is the magnetic recording head which we shall consider separately.

Piezo-electric effects have been known since Kelvin postulated permanent polarisation of certain crystals in 1878 and the Curie brothers investigated piezo-electric effects in crystals in 1880. Woldenar Voigt laid down much of the basis of crystal physics long before a practicable application was found. Perhaps the earliest example was the underwater sound detector of Paul Langevin during the First World War-though even this was preceded by a British patent taken out by L. F. Richardson in 1912. The real growth of piezo-electric development occurred in the "Twenties" when Sawer and Tower founded the Brush Crystal Corporation.

Before 1939 only quartz and Rochelle salt were used as crystal resonators and gramophone pick-up units. But the piezo-electric effect can be demonstrated in 20 of the 32 crystal classes. The underlying principle is similar but applications depend on the limiting factors of cut-off frequency temperature sensitivity and the humidity effect. For commercial transducers types are limited to three kinds of crystal. These are: (a) Ammonium dihydrogen phosphate (A.D.P.) (b) Sodium potassium tartrate
(Rochelle salt) and (c) Polycrystalline barium titanate (ceramics).
A.D.P. has a higher volume sensitivity than Rochelle salt but a very low dielectric constant. It is responsive to humidity changes but is fairly stable with temperature changes. The great drawback is a high and variable conductance which gives it a high cut-off frequency. (At frequencies below resonance crystal transducers behave as generators in series with a capacitor terminated with a resistance). Matching must be into an impedance of $10 \mathrm{M} \Omega$ and above and this limits its usefulness.

Rochelle salt is humidity-conscious and varies widely in its characteristics because of the temperature sensitivity. Much depends on the conditions under which transducers are made. Excess moisture must be excluded. But when units such as microphones are hermetically sealed, Rochelle salt units are effective, and have the advantage of cheapness. It has the highest electro-mechanical coupling coefficient of any commercially-available piezoelectric material. The electrical load into which it matches must be high compared with the reactance of the crystal. More about this, and the barium titanate characteristics when we discuss pick-ups.

## Microphones

Cheapest and most popular microphones that need be considered are the crystal types, with Rochelle salt of quartz crystals, acted upon by mechanical movement as the air fluxes a diaphragm which stresses the crystal and produces an e.m.f. Thus, the crystal microphone is pressure-operated.

It has different methods of construction, the simplest type being a dimorph, or two slices as shown in Fig. 1(d), made this way to give extra voltage, the e.m.f., from each half adding. Upper frequency response is little more than $6 \mathrm{kc} / \mathrm{s}$, making these units applicable for speech, but unsatisfactory for musical reproduction. By sealing two dimorphs in a cavity, leaving an airspace, the mechanical connection is eliminated and frequency response may be extended to $10 \mathrm{kc} / \mathrm{s}$ or more. It is quite possible to improve upon this by careful selection of crystal and construction, but costs rise disproportionately.

Microphone sensitivity can be plotted as the voltage obtained from a constant sound source at various angles around the unit. This "polar diagram" is drawn in one plane, but extends all round, and the theoretical polar diagram for a pressure-operated microphone is as Fig. 1(a). This is modified by the shape of the microphone housing, and, in practice
by the frequency of the sound source.
At low frequencies the air strikes the one-sided diaphragm at more or less equal strength whether approaching from the front or back but higher frequencies do not permit the pressure changethere is not enough time for the diaphragm to respond - and the response becomes more and more unidirectional. So the circle of Fig. I(a) becomes elliptical as the frequency of the sound source rises.

A crystal microphone is effectively a generator in series with a capacitor when matched into its appropriate load. The reactance of the capacitance varies inversely with the frequency and C forms a potential divider with the load Z. At lower frequencies $C$ is significant the reactance is greater and more signal is lost across C . Cable capacitance must also be taken into account and as typical values might be 150 pF per foot for the response to be no more than 3 dB down at $40 \mathrm{c} / \mathrm{s}$ matching impedance should be $8 \mathrm{M} \Omega$ or greater with a microphone capacitance of 500 pF . So bass response is lost if the load impedance is not high enough.

High impedance lines are prone to hum pick-up and this is another limiting factor in the use of crystal microphones. Coupling into the low impedance of a "conventional" transistor circuit is poor so bootstrap inputs with directly coupled pairs of

## Moving coil microphones

Perhaps the most popular type of general purpose microphone is the moving coil unit whose general construction follows the sectional diagram Fig. 1(b). This is also a pressure-operated device essentially robust and capable of a very good frequency response when well designed. Ideally the polar diagram is as (a), the microphone being omnidirectional at low frequencies and more directional from front and rear as the frequency of the sound source increases. Modifications to the housing and types of baffle permit the moving coil microphone to achieve a cardioid polar diagram as in Fig. 1(c). By fitting an acoustic labyrinth to the microphone, it can also be made into a very effective unidirectional unit.
The principle is simple. A coil is wound on a cylindrical former to which a cone-shaped diaphragm is attached. The former sits in the annular gap of a magnet. Air movement causes the cone to move, cutting lines of force with the coil, into which an e.m.f. is induced.

Frequency response of a good MC microphone can extend from $40 \mathrm{c} / \mathrm{s}$ to $15 \mathrm{kc} / \mathrm{s}$, and resonances due to the mechanical system can be reduced to quite small proportions--although for all but the best
 types, these are still limiting factors. The impedance is low, and the output voltage is lower than that obtained from a crystal microphone, and for this reason a transformer is often built into the microphone housing, as in Fig. 2. The output is raised and the matching impedance may be $200 \Omega, 600 \Omega$ or $50 \mathrm{k} \Omega$ in typical devices. Very often higher quality microphones have an inbuilt switch that allows the user to select low, medium or high impedance matching.

Other factors that must be considered when choosing a microphone are the sensitivity and the front-to-back ratio. The latter specification is given in decibels, a typical figure of 15 to 20 being quoted for MC units, with values from unity to 20 or more for ribbon and capacitor microphones, specially constructed for directional responses.

Sensitivity may be quoted in microvolts per micro-Bar ( $\mu \mathrm{V} /$ $\mu$ Bar), the Bar being a unit of air pressure, decibels relative to a stated voltage reference, or simply in microvolts. The reference level is made to equate with 0 dB , and can be expressed as the voltage produced at a certain sound pressure, or the output across a load of known impedance. Thus 0 dB
transistors have been developed to accommodate high impedance sources.
Carbon microphones are also pressure-operated devices but their inherent noise level and the restricted frequency response make these applicable to limited uses such as telephones and they need not be considered here.

(d)

Fig. 1: (a) Polar diagram of pressure-operated microphone. This is "flat" section: microphone is senstitive "upwards and downwards" as well as in forward plane. (b) Sectional diagram of moving coil transducer. The equalising tube is employed to prevent back pressure affecting the performance. (c) Cardioid polar diagram, available for many moving coil microphones. (d) Crystal units cut from the basic crystal in the manner shown. The hard line shows the shape of the original crystal, the line-dotted shape is the basic plate, with force lines as shown, X,Y and Z, and the dotted slice is the final cut from the crystal to give bimorph properties which produce e.m.f. from mechanical stresses. (e) Figure-of-eight polar diagram, typical response of ribbon microphone. (f) sectional view of basic ribbon construction.


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fact, everything you need to build a compact speaker system.

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$600 \Omega$ (which works out to 775 millivolts). So the same microphone with -60 dB rating would give 0.000775 volts or 0.775 mV . Where the output is in volts or fractions of a volt this will generally be the standard implied. To work out the voltage from the "minus dB " specification we use the following formula:

$$
E_{1}=\frac{E_{2}}{\text { antilog } d B / 20} \text { (because } d B=20 \log \frac{E_{2}}{E_{1}} \text { ) }
$$

$$
\text { thus } E_{1}=\frac{0.775}{\text { antilog } 60 / 20}=\frac{0.775}{1,000} \text { volts }
$$

## Ribbon Microphones

The conventional ribbon microphone is a pressure gradient operated device. A very thin corrugated ribbon usually of aluminium is suspended between the poles of a magnet and protected by acoustically transparent grilles. Movement of the ribbon is caused by the difference in pressure between the front and the back and the polar diagram is thus a figure-ofeight as in Fig. 1(e). This makes the ribbon microphone especially useful for stage and studio work where sounds from the side directions can be virtually excluded. Actors can come in and out of sound focus by merely moving a foot or two sideways!

By fitting acoustic pads the bi-directional characteristic can be modified to a cardioid response. The very good frequency response and the sensitivity to transients (short-term sounds) makes these microphones suitable for much high fidelity work. But the construction renders these instruments more vulnerable to damage and they are not suitable for outdoor work unless well shielded from the wind.


Fig. 2: Basic construction and circuit of typical moving coil microphone with inbuilt transformer.

Because of the excellent frequency response of the ribbon microphone bass will be accentuated if the sound source is too close. Impedance and output voltage are low for the basic unit and it is common practice to build a transformer within the shell of the microphone. Units with a terminal impedance of $30-50 \Omega$ are common requiring further transformer matching to most amplifier inputs.

## Capacitor Microphones

Again a pressure-operated type as is the moving coil, the capacitor (or condenser) microphone can be made to very high quality specifications with a frequency response extending well beyond anything required for normal audio work. A diaphragm forms one plate of a capacitor, a rigid piece of metal the other. A polarising voltage is required, and as $\mathrm{V}=\mathrm{QC}, \mathrm{Q}$ being unable to change rapidly, the capacitance is very nearly proportional to the voltage, and movement of the diaphragm gives a voltage variation in time with the sound. A flat response of $20 \mathrm{c} / \mathrm{s}$ up to $15 \mathrm{kc} / \mathrm{s}$ and more can be obtained with quite normal capacitor microphones although these units are not cheap. Response can be omni-directional, bi-directional or cardioid, depending on construction.

A three-plate capacitor can be used, with a diaphragm on either side of a perforated plate. As the diaphragms are mechanically coupled by the air between them, they move simultaneously. With one diaphragm polarised, a combination of pressure and pressure gradient operation is obtained, and a cardioid response. Polarising the second diaphragm to aid the first, produces and omni-directional polar diagram, and when the two oppose, a figure-ofeight polar diagram. Impedance can be low or very high, and output is comparable with MC units.

## Pick-up Cartridges

Some of the widest price variations in the high fidelity market are encountered as we take a look at the range of gramophone pick-up cartridges available. These can be of the three first types of transducer we considered: (a) piezo-electric, (b) moving coil or (c) moving armature (-variable reluctance). The cheapest types are generally in bracket (a) and these can be further sub-divided into crystal and ceramic units, with the latter giving less output but capable of high quality.

Rochelle salt crystals are of the general dimensions $0.5 \times 0.25 \times 0.03 \mathrm{in}$. and ceramic' (barium titanate) bimorphs $0.7 \times 0.1 \times 0.03 \mathrm{in}$. and a practical stylus length may be as little as 1 cm . Normal capacity is between 500 and 1500 pF . Compliance (freedom to move) of the crystal is lower than required for good tracking and the construction demands very precise engineering the actual crystal being supported on resilient pads within the cartridge body.

The resonant frequency of such a system is 1
given by the formula: $\mathrm{f}_{\mathrm{r}}=\overline{2 \pi \sqrt{ }(\mathrm{M} . \mathrm{Cm})}$
where $M$ is the effective mass or the moment of inertia and Cm the effective compliance. The resonant frequency is lower for cantilever than for torsional units.


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The output voltage from an ideal pick-up cartridge would be proportional to the velocity of the stylus. With sinusoidal recording movement of constant velocity the peak amplitude of the recorded groove (and thus peak displacement of stylus) would be inversely proportional to frequency. So if low frequency signals of large amplitude are to be recorded the groove spacing must be large. As this would reduce playing time per disc and impose serious tracking problems-about which more later -the ideal velocity characteristic is modified. Maximum amplitude of stylus travel at low frequencies is limited and some bass boost is necessary in the amplifier to compensate. This is one reason for equalisation which differs according to the type of cartridge in use.

Another factor is the low power content of the highest frequencies (small amount of actual stylus travel) which can become comparable to the noise signal caused by mechanical discrepancies. To overcome this the treble is pre-emphasised during recording. Acceleration of the stylus due to preemphasis is proportional to frequency. This accentuates the tracking problem still more with great differences between grooves at the outer edge and toward the centre of the disc. So the recording characteristic has to take account of this factor.
For similar reasons the recording characteristic and this the equalisation that must be applied varies according to the speed of the record. Standardisation of equalising curves was proposed at the CCIR conference in Philadelphia in 1954 and the Recording Industry Association (RIA) conforms using standard sizes for microgroove dimensions and stylus radii, as follows: Coarse groove 78 r.p.m., 0.006in max groove width with 0.001 stylus radius and a groove angle of 80 to 93 degrees (taking the groove as a V -section). Fine Groove, 33 and 45 r.p.m., similar groove angle, with 0.002 in groove width at top and radius of 0.003 in . Stylus radius is then limited to $0.002-0.003 \mathrm{in}$ for coarse and 0.0008 to 0.001 in for fine groove replay.
Because crystal and ceramic cartridges are capacitative by nature the matching is into a high impedance, and equalisation is necessary. To obtain what is known as a velocity characteristic, the ceramic cartridges can be matched into a lower impedance, dispensing with subsequent equalisation. There is one snag which is seldom mentioned, however, and this is the capacitance versus temperature anomaly of crystal designs. Low frequency correction will be a function of temperature and cartridges thus forced into the velocity mode can exhibit wide differences in performance according to the ambient temperature and humidity.

## Magnetic Cartridges

Before discussing other factors that affect the playing of discs, and design of cartridges, we should take a brief look at the alternative types of transducer, the moving coil and variable reluctance cartridges.

These are generally of lower output but much better frequency response, with resonances that affect replay less, and capable of lighter playing weight, which reduces many mechanical drawbacks. They are thus more expensive and better engineered, and within the type there are very wide differences in structure and quality-and price.


SOIDP
Many of the early pick-up cartridges were of the moving-iron type, doing their job quite successfully, almost by the brute force assistance of their weight, which kept the stylus in the groove despite its low compliance. There is as much difference between the modern variable reluctance types and these older versions as between a racehorse and a Suffolk Punch. Modern pick-ups are capable of tracking at one or two grammes and have compliances of 15 or 20 times $10-6$ dynes $/ \mathrm{cm}^{2}$.

Of the modern types the variable reluctance cartridge is currently leading the quality field. There are two main designs the single-ended reed system and the balanced armature cartridge. The basic principle is of an e.m.f. generated by the change in flux through a fixed coil. The magnetic circuit is energised by a magnet in which the reluctance (the magnetic equivalent of electrical resistance) can be varied by displacement of an armature.

The e.m.f. generated is proportional to the number of turns on the coil the magnetic force the area of the polepiece and the velocity of the armature and inversely proportional to the spacing between armature and pole. Armature velocity is governed by the driving force i.e. the movement of the stylus imparted by the varying groove of the rotating disc. But mechanical impedance and the reflected electrical impedance due to normal transducer action also affect the velocity and thus the output. An improved method of variable reluctance design is the balanced armature configuration.


Fig. 3: Sectional diagram of balanced armature transducer, with (b) electrical analogy of the magnetic circuit. This design is widely used for variable reluctance cartridges. $F$ denotes the mechanical force employed to move the armature.

If the armature is symmetrically disposed between four gaps as in Fig. 3 (a) a balanced magnetic bridge is obtained. Less restriction on the armature results. The electrical analogue of the magnetic circuit makes this clear. Figures 4 and 5 are also of interest to the user with the unique Decca cartridge an example of a different approach to the problem of non-linearity inherent in the variable reluctance design.


Fig. 4: Simplified diagram of typical moving magnet type of pickup.

With a stereo cartridge two sensing coils are needed the groove walls being cut so that the stylus follows lateral variations for the left channel and vertical variations for the right channel. (This is necessarily a simplification-the actual shaping of the curve is a combination of the two modulations the principle of addition and subtraction of forces being used to sort out the channel separation.) In


Fig. 5: Principle of the Decca ffss Mk 4 cartridge.
the ffss pick-up, the sensing coils are at right angles, one coil being lateral and two coils vertical. Sum and difference voltages are thus produced. An inverted " $L$ " armature is employed, with vertical compliance provided by the fiexibility at the top of the " $L$ " and lateral compliance by the freedom of the " $L$ " to pivot. The armature is made of magnetic alloy, with low internal damping. This keeps the resonance of the mechanical circuit below $10 \mathrm{c} / \mathrm{s}$, and this is further controlled by the pickup arm design.

Tip mass resonance for the low tip mass of 1 mgm is kept high, about $30 \mathrm{kc} / \mathrm{s}$, and frequency response is well defined. In the later versions, even this aspect has been considered and damping is applied at the armature clamp. The vertical tracking angle is at the now accepted standard of 15 degrees, which means that the stylus 'prods forward' from the vertical by this angle when seated correctly in the groove and with the arm correctly set. Arm and head setting is vital for reduction of distortion and protection of records. More will be said about this in the following section on measurement and testing.

## Pick-up Terms

Some of the language of the hi-fi enthusiast needs explanation to those entering the field. In explaining the terms, we may be able to throw some light also upon the reasons for certain parameters.
Compliance we have already noted is the freedom of the stylus to move. The stylus is moved by the shaping of the groove, so a force is exerted upon it. To keep it in the groove, a sufficient tracking weight is needed. The two things go hand in hand; the lower the compliance, the greater the tracking weight that will be needed. Similarly, the greater the modulation amplitude (and hence the stylus displacement), the more tracking weight will be needed. For any given set of conditions, there will be an optimum weight. Reducing this arbitrarily can be a mistake, allowing the stylus to "jump" in the groove and causing damage to the groove walls.

Compliance is measured as the distance of movement caused by a force. The unit force is 1 dyne, which is approximately a thousandth of a gram. Distance is measured in millionths of centimetres. So the compliance will be so many times $10^{-6} \mathrm{~cm} /$ dyne. For stereo cartridges both vertical and lateral compliance must be stated. The higher the compliance figure, the "better" the cartridge, other things being equal (which, of course, they never are).

Stylus mass or tip mass must also be considered. This should not be confused with playing or tracking weight, which may be 1 or 2 grams, or even more. Tip mass should be as low as possible, and will be measured in milligrams. Many makers do not publish this specification, but it matters immensely, and, in the opinion of some authorities, is the most vital single factor in pick-up design.

In general, it can be said that for any given tip mass in milligrammes, the tracking force or playing weight is a similar figure in grammes. The cartridge cannot be considered alone, and the friction of the arm must also be taken into account, so that the actual playing weight will in many cases be more than this figure.

The three factors already mentioned are governed by manufacturing methods, and for any given set of circumstances, i.e. modulation level, speed, groove disposition at beginning or end of record, etc., these will be slightly modified. The end result will be a compromise. Too great a departure from the optimum will result in groove deformation, and this can happen on the first playing and is irrevocable. Hence, it is a mistake to lend a disc to someone whose equipment may be set to different playing weight, or whose cartridge has less compliance or pick-up arm more stiffness than one's own.

At higher than optimum weights, the stylus pressure generates enough heat to temporarily liquify the portion of groove wall over which it passes, and as it cools, the wall takes on a distorted shape, which will always be there. The elastic limit of the record material is a fixed amount, and is a factor of playing weight relative to tip size, measured in lbs./ sq. in. The lower the stylus contact area, the lower the tracking force that should be used.

Dust and cutting swarf that may remain in a groove also complicate matters, and there is the fallacy that some records, which may have inherent distortion, sound better when played with a greater tracking weight. The effect is caused simply by the brute force blanking of transients. Very often, the true answer would be more meticulous care of discs. Many cleaning agents can be used, and even a simple disc preener or swab will prevent much groove deformation, if regularly used.
Frequency response is determined largely by the tip mass. About $10 \mathrm{kc} / \mathrm{s}$ may be available from a 4 milligram stylus and this response can be doubled with a 1 mg stylus. The lower frequency limit is determined by the resonance between the pickup's compliance and the combined mass of cartridge and arm. Other undesirable resonances can occur within the frequency range to upset the "flatness" of the response in badly designed systems. Damping can improve matters and distributed mass is a designer's way of reducing unwanted resonances. Compliance, tracking weight, tip mass and the overall damping are taken into account, and the reaction force at the stylus can be measured at various frequencies throughout the range to produce an impedance curve, which can often be a bit of an eye-opener.
Temperature can alter tracking weight requirements. In a very cold room a crystal or ceramic cartridge may need greater tracking weight. In some cases, a decoupled stylus is used to take advantage of temperature and other relevant changes.

Sensitivity is measured in r.m.s. volts for a certain stylus velocity. Thus, we find specified: $\mathrm{XmV} / \mathrm{cm} / \mathrm{sec}$. This means that for each centimetre per second of stylus velocity there will be an output of X millivolts, and explains why a cartridge of specifications say $3 \mathrm{mV} / \mathrm{cm} / \mathrm{sec}$ will match a preamplifier and fully load it when the sensitivity, say 7.5 mV , of the amplifier input is two or three times the basic output of the cartridge
Normal recorded velocity on dises is taken to be $5 \mathrm{~cm} / \mathrm{sec}$; so a pickup with 1 mV per $\mathrm{cm} / \mathrm{sec}$ will be expected to give an output and match a preamplifier with the sensitivity of 5 mV . Output is not the most important criterion. As a general rule, the better the quality of the pick-up, the less its output voltage.

Trackability is the degree of correctness with which the stylus can trace out the groove, in exact relation to the cutter which produced it. Cutters are not pivoted like pick-up arms, but move laterally across the disc, keeping the cutting head always at right angles to the groove. The inward motion of the arm should thus lie as nearly as possible along a radial line from edge to centre of disc, with the pick-up mounted so that the stresses due to the rotation are compensated. There are various adjustments that can be made, differing between arm and head makers, as can be seen from the accompanying

illustration. In the next part we shall look at some of the important adjustments and methods of checking tracking. In general, the longer the arm, the less the curvative of the pick-up motion and the better the tracking, but this also depends on arm shape, balance, pivot stiffness and several related factors.

Tracking error is affected by the size of the stylus, as is scanning loss. This is the loss of high frequencies because of the relationship of stylus size to the actual wavelengths of the recorded sounds. It is more evident at the centre of the record, where physical wavelengths are smaller. Scanning loss is reduced by the use of stylii with smaller radii.

Stylus sizes (radii) for microgroove records may be 1 mil. ( $25 \mu$ ) mono and $\frac{1}{2}$ mil ( $1 \cdot 3 \mu$ ) for stereo. 'Compatible' stylii designed to replay either mono or stereo with the stereo channels of the cartridge paralleled for mono operation are generally 0.7 mil $(18 \mu)$. The use of the smaller stylus can help reduce scanning loss and tracing distortion but only if the arm will track at less than about 3 grammes. On older discs and some earlier stereo records bottoming may occur with the stylus sitting too deeply in the groove. This can be avoided to some extent by using bi-radial or elliptical stylii. These have a major radius of 0.7 to 0.8 mil across the groove with the edges contacting the groove walls having the minor radius of 0.2 to 0.3 mil. Elliptical stylii can help reduce distortion effects due to overmodulation at high frequencies.


Fig. 6: Various adjustments on pick-up arm can be seen in this view of the popular Lab 80, with the recommended Shure M75 cartridge.
Pinch effect is a form of tracing distortion more noticeable on mono records occurring where the groove walls move in phase. The groove will be narrower in the middle of the sideways travel of the cutter. This presents a narrower groove to a spherical stylus twice every cycle causing it to ride up and down twice for every lateral cycle. Using a stereo pick-up to replay mono records can produce second harmonic distortion unless the two channels are paralleled to cancel out the vertical motion's effect.

Needle talk can also result from the unwanted movement of the stylus in the groove. This is the noise of the pick-up playing when the amplifier is


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turned down and can be caused by the reaction of the disc to the movement of the stylus. Excessive needle talk at high frequencies is generally caused by high tip mass and at low frequencies by low compliance.

Surface noise can be misleading. With modern vinyl records played at less than 3 gm weight noise due to the frictional effect of the stylus in the groove should certainly be 70 dB or so below the normal signal and generally inaudible. Some discs are notoriously 'noisy'. But, less frequently realised and of equal importance, the structure of pick-up arm and playing deck can add much to the relaying and amplification of so-called surface noise. Thin-walled pick-up arms and lightweight plates add much to the noise problem. Unwanted resonances complicate matters still further.

Decoupled stylus is a term obtaining more recognition. A flexible member couples the stylus arm to the static parts of the pickup, acting as a kind of shock absorber. By further decoupling the rondel (stylus shank) from the stylus arm, an even lower tip mass can be produced. The problem is to decouple in this way and still avoid the loss of high frequencies, and of stereo separation.

Anti-skating devices are used to compensate for the natural inward pull of the pick-up arm, and take the form of bias correctors. More will be said about these in the section devoted to measurement and performance checking. Low pivot friction is needed to counteract stiffness, and this allows the forces exerted on the stylus to become more evident in one direction and to change according to the angle of the arm. Bias correctors thus act to compensate for a changing side-thrust. A longer arm can reduce sidethrust caused by the 'drag' of the moving disc.


Fig. 7: Pritchard pick-up system employs thread-and-weight compensator device to equalise the pressure on both walls of the groove. Ball and cup bearings in gimbals reduce vertical and lateral friction. Playing weight is adjusted by patented 'flyaway' calibrator.

Rumble can originate in the action of the turntable. It will be reduced by good bearings, a heavy turntable and generally good engineering. But some rumble is inevitable: this does not mean it should be heard. It can be further reduced below audibility by good pickup damping and good arm design, and by the solid mounting of the playing deck so that
physical transmission of sound back to the stylus is avoided. Rumble below 30c/s will not normally be heard, and if the lower resonance of the pick-up system can be arranged to be just above this frequency, the falling response below this resonance will help cancel out rumble. But adequate damping is needed to cancel out the resonance, and this is a question of overall design.

Vertical response of the pick-up is most concerned with rumble sensitivity. If the vertical cut-off frequency is higher than the lateral, some improvement can be gained. Moreover, if the vertical inertia of the arm is greater than the lateral inertia, even more protection against rumble is given. Pick-ups often have a vertical to lateral compliance ratio of $1: 3$, giving a lateral response flat to about $25 \mathrm{c} / \mathrm{s}$, a stereo response to about $30 \mathrm{c} / \mathrm{s}$ and a vertical response that cuts off at about $45 \mathrm{c} / \mathrm{s}$.

## Loudspeakers

Perhaps the best-known form of transducer is the loudspeaker, and of the various types, the moving coil design is by far the most popular. But this is by no means the only type of loudspeaker design to achieve success. Gilbert Briggs, in his immensely readable book(s) on loudspeakers cites many examples of basic design beginning with the original moving iron and balanced armature types and the horn attached to a telephone earpiece which was surely the earliest attempt at electro-acoustic amplification. Of the types one or two have been passed over lightly, but are beginning to come back into fashion with improvements in materials and development. Among these may be mentioned the electrostatic speaker, the ribbon speaker and the Ionophone design.

## Moving-Coil Speakers

Figure 8 shows a cross-sectional view of a typical twin-cone loudspeaker of good quality. Although possessing several special features, it embodies the fundamentals of moving-coil design. The coil is attached to a cone-section diaphragm, and is suspended in the annular gap of a fixed, permanent magnet. Currents through the coil cause flux-cutting of the magnetic lines of force and the cone moves, driving a body of air before it to produce the sound. This is the inverse to the principle of the moving coil microphone, which we studied earlier in this section.

Modern loudspeaker magnets are of ceramic material, with a high flux density. The type illustrated in Fig. 8 has a flux density of 16,500 gauss and a total flux of 185,000 maxwells, using a $1 \frac{3}{4}$ in. diameter voice coil of aluminium construction (for weight reduction) and a heavy diecast chassis. Handling 20 (British) watts of power, its frequency range is $30-16,000 \mathrm{c} / \mathrm{s}$ and its fundamental resonance is at $35 \mathrm{c} / \mathrm{s}$.

One feature of this design is the plastic suspension of the cone to achieve greater flexibility. There are various points of view about cone suspension and cone stiffness, but in general we must remember that the cone itself should not contribute any distortion due to resonances. Several factors affect performance. Size is the first consideration: all other things being equal, the larger the cone, the better the bass performance. A greater diameter enables a greater mass of air to be moved and more power can be developed. But this depends also upon the cone excursion.

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Japan's leading producer of transistorised High Fidelity equipment has a "small" name-but a name synonymous throughout the world with forward thinking design and high quality manufacture. TRIO transistor circuitry is highly developed and produces reproduction characteristics comparable to the valve equipment for so long used as the yardstick of quality-thus offering complete satisfaction to the ear of even the most critical Audiophile.


The range of Stereo Amplifiers and Tuner Amplifiers covers units with power ratings of $12+12$ to $70+70$ watts RMS per channel and includes the only multi-channel (three amplifiers per side) Stereo Amplifier at present available in the world-the "Supreme 1" with 142 watts total RMS power. Input sensitivities of only 2 mV on pick-up and tape inputs, suitable for all the new generation of low output cartridges and tape heads, are standard on all models.
Let TRIO speak for itself-ask for a demonstration at your leading High Fidelity Showroom, it will be a rewarding experience and by buying TRIO you will almost certainly pay less-and get morel


Illustrated above: the new outstanding value TK-150T $13.5+13.5$ watts Stereo Amplifier-recommended retail price $£ 35.0 .0$, shown with HS-3 Stereo Headphones-recommended retailprice $£ 5.15 .0$. Right-top to bottom: compact model TK-250T $20+20$ watts Stereo Amplifier-recommended retailprice £49.10.0; TK-350T AM-FM Stereo Tuner-recommended retailprice $£ 56.0 .0$; TK-40LT AM-FM (MW/LW) Stereo Tuner/ $16+16$ watts Stereo Amplifier-recommended retail price £98.0.0.

Illustrated leaflets giving full technical details of these and other TRIO models are available on request with the name and address of your vour nearest appointed TRIO dealer.



Ever had trouble locating a particular piece for your project? Ever wasted time thumbing through confusing price lists? Ever been
foot-weary and frustrated tramping round the shops?
Sigh no more. Just sink into an armchair and enjoy life with a Home Radio Catalogue!
plus 3'p\&p.
A bargain at $7^{\prime} 6$ Pick your parts. Grab your pen. Make for a letter-box. Your chosen items will be with you almost before you can get back to that armchair!

[^4]Shape and angle of cone are other important factors. Circular cones are easier to make than other shapes, but elliptical and (practically) square cones are favoured by many manufacturers.

The elliptical cone is a method of obtaining the largest possible cone area in a given space. The cross-sectional shape may be straight or curved, and the use of straight cones can cause 'break-up' effects if the suspension is not specially designed to guard against this phenomenon. Cone break-up is caused by low, damped resonances in the cone itself. This produces a rise in output at mid-frequencies.
A flared cone can reduce this effect, but the production of sub-harmonics is then a danger. Intermodulation distortion can be caused, and again, good suspension design-and enclosure design-is needed to obtain the best results.
The angle of the cone affects directionality. Shallow cones are less directional than deep ones, but are weaker when fully powered at the lower frequencies. Ribs and corrugations in the cone can modify the above characteristics, and various means are favoured by different makers.

The weight of the cone, dependent on the material and its thickness (and rigidity), help determine the fundamental resonance frequency. The heavier the cone, the lower the resonance, but the lower the efficiency. Light cones with highly compliant suspension systems can lower the resonance, but voice coil centring is more difficult and cone rigidity harder to achieve. For this reason, materials of greater inherent stiffness (such as light metals) have been used in some designs. These are more common for high frequency units. Great care is needed to avoid peaky responses due to hard cones. As most of the high frequencies are radiated from the cone centre region, a compromise is possible, with the central portion stiff and the outer region softer.

Compliance is important to obtain a low resonance. This can be improved in different ways. Corrugations of the outer edges of the cone, treatment of the outer section of the cone with viscous damping agents and soft surrounds of various materials that have included leather, goatskin, velvet and expanded polyurethane foam, are all methods of obtaining low cone resonance.

As important as the cone and its suspension is the way it is enclosed. Cabinet design is a subject too large for the terms of reference of this supplement, but a few words on general principles may be appropriate. In particular, it should be remembered that the cabinet-more properly, the enclosure-should not of itself contribute to the distortion level, but must nevertheless be considered with the loudspeaker unit as a combined radiator of sound.

The criterion for one popular design was undoubtedly reduction of size for the same frequency coverage and power output as a larger unit could produce. To this end a light and rigid metal diaphragm of only 4 in . diameter was employed, but with special suspension and speech coil to magnet relationship that enabled the maker to form a module capable of reproducing a claimed 20 to $20,000 \mathrm{c} / \mathrm{s}$ at low distortion, with power distribution afforded by combining the modules. Each unit handles 12 W r.m.s.

Quite simple cabinets were recommended by the makers, with a single rectangular opening in the front and a port and tunnel arrangement on the

inside surface. Lining of the cabinet requires complete acoustic filling. The material recommended for this model is bonded acetate fibre. Not all designs use the same filling, and care must be taken when fitting pads and filters that the right materials are chosen. Joints of the cabinet are made airtight, screwed and glued, with foam plastic sealing between the module and the top of the tunnel. A further example of acoustic wadding is given in Fig. 9 , where pads of fibrous material are hung vertically in the space behind the speaker unit.

Resin-bonded Fibreglass is used by several authorities for sound absorption. The principle behind absorbents is that sound from the radiator should be allowed only in a forward direction, but it must be free to travel, so acoustically "dead" material is needed in an enclosed space to prevent the sound waves "bouncing around" the cabinet. When they are free to do this, standing waves are built up, the resonances due to the cabinet shape give enhanced sound at some frequencies and act as deadeners to others and completely upset the frequency response of the system. Moreover, the effect of the sound wave reflected from the cabinet inside to the cone of the speaker is to damp the piston action and reduce overall power. Cabinet design is an advanced study: wherever possible, recognised designs should be followed, cabinet and loudspeaker unit being considered as a whole.
Baffles are the plates on which loudspeaker units are mounted. They must be flat and rigid, and the material has to be chosen to have least effect on


Fig. 8: Cross-section of 12 -inch version of the widely used Goodman Axiom loudspeaker (twin-cone, single-unt).

the sound. Wood is still reckoned to be the best material, despite experiments with many others, including, marble, aluminium and slate. For optimum results with a given weight and cost, two sheets of plywood with sand between are recommended by a leading maker.

This is for a flat baffle, whose response tests show a fall of 6 dB per octave at frequencies below the critical point determined by size. To reproduce $C_{3}$, for example, the bottom $C$ on the piano keyboard, whose frequency is $32.7 \mathrm{c} / \mathrm{s}$ and wavelength 35 ft ., the minimum diameter baffle suspended in free air would have to be 17 feet-rather large for the living room Mounting a loudspeaker unit in the wall of a room is one way of achieving a near infinite baffle.

So-called infinite baffle enclosures make use of a slightly different principle. By totally enclosing the loudspeaker, making the box airtight, and filling it with wadding, all back radiation is prevented and in theory all the radiation of sound comes from the


Fig. 9: Cabinet resonances are reduced (in theory, eliminated!) by inserts of special acoustic wadding in the sealed cabinet, as shown.
er and of rigid construction. When added to a bass unit with a properly designed crossover unit, a good frequency coverage can be produced. The bass unit is relieved of the necessity of handling the rapid excursions of the high frequencies and less cone "break-up" will be encountered.

A tweeter will be added to an existing system in several different ways, as shown in Fig. 10. The simplest method, (a) is a high-pass filter, giving a crossover at $3-5 \mathrm{kc} / \mathrm{s}$. The capacitor should be a good paper type, but electrolytics can be used after being "d.c. aged". This is done by applying a voltage of about a quarter to a fifth the rated d.c. working voltage, first in one direction, then the other, to stabilise the electrolyte. But when modern metallised capacitors are available so cheaply, this hardly seems worth while.

Additional types of loudspeaker of great popularity, but higher cost, include the "tweeter with no moving parts", based on the original Ionophone. An r.f. oscillator ionises a column of air which is modulated by the sound and produced via a horn. Very clean "top" is possible when this unit is properly matched. front. But there are various parameters to be considered; cone mass, suspension stiffness, unit compliance, radiation resistance and air load reactance all modify the theory.

Surprisingly good results can be obtained from "shoe-box" sized loudspeaker enclosures built on the infinite baffle principle. More power is needed to drive these units, and bass cut-off is generally higher than with a larger reflex cabinet, or alternative "open" design. However, where the overall system has deficiencies in the lower frequency region, this can be a good thing!

High frequency speakers, or tweeters, are directional and handle less power, so can be much small-

The electrostatic loudspeaker has long been the pet of one leading company, and is capable of remarkably good wide-range reproduction. It is often difficult to match, especially to a transistorised amplifier, but well repays the extra trouble taken, and the slightly higher cost.


## No. 21

## Redbridge

 AMATEUR RADIO SOCIETYSOMEONE said something to the effect that great oak trees grow from little acorns which is a pretty good description of how the Redbridge Amateur Radio Society have progressed.
In July 1967 four radio "acorns" met-G3JHY, G3JTS, G3ODS and an s.w.l. called Bill. News, like r.f., travels fast, and by the following month the number had risen to eight. The increase in membership continued but unfortunately as the membership swelled, the size of G3JTS's living room remained constant and it was decided that a larger premises was a dire need if the meetings were to continue.

After many reconnaissance delegations had prowled the district and returned empty-handed, G3ODS managed to locate a suitable room in a local church premises at a nominal rental, and the Redbridge Amateur Radio Society had found a home. The rental for the room is recouped by prising a "bob" from each member's purse as he enters the door. There are no trading stamps, but the shilling so invested does entitle the donor to a cup of tea and biscuits at half-time. If you happen to be under sixteen, it only costs sixpence.

Meetings have continued at this QTH ever since and take place on the first and third Mondays of each month except Bank Holidays. Topics vary from the time-honoured ragchews to demonstrations, lectures, junk sales and open discussions etc. The club hopes to be able to install some gear including transmitting equipment in due course so that licensed members can operate from the club while s.w.l.'s can see for themselves just how it's done.

Short wave listeners and beginners are very welcome at Redbridge, in fact quite a large number are in regular attendance. Every encouragement is given to those wanting to get an amateur licence. On Thursday evenings, chirps and tweets can be heard emanating from the QTH of G3JTS. This is a regular morse practice session and, at the time of writing, there are eight keen types in attendance. G3JTS calls a halt for a cuppa around 20502100 hrs (clock time) and then it's heads down again for more c.w. practice.

Anyone interested in radio, whether licensed or not, is very welcome at Redbridge. The club is also very pleased to welcome visitors. It doesn't matter what you do for a living, who you are or what knowledge you might or might not possess, the only qualification to join the club is an interest in radio. The club has purposely fixed its meeting nights in order not to clash with other local groups at Chadwell Heath and East Ham.

Meetings are held at the Presbyterian Church, corner of Oakfield and Albert Road, Ilford, Essex (near Ilford Town Hall). Anyone is welcome on Club nights, or you can contact Bob (G3JHY) at 70 Kensington Avenue, Manor Park, E12, for further particulars.

By-the-way, "Not knowing much about radio" or "being a beginner" are not accepted as being a good enough excuse for not joining the club, so how about looking us up? Anyway, where else could you get such a lovely cuppa and biscuits for a bob?

[^5]

## THE BROADCAST BANDS

by CHRISTOPHER DANPURE

T1HE month of October besides being the last month of the Spring/Autumn schedules is also a month when great changes occur in propagation conditions. During September there was not much change in these propagation conditions, but when October starts you will find the higher frequency bands will close considerably earlier than in September and open much later, and stations over short distances will start to skip greater distances earlier in the evening. Generally there will not be much difference between the October and November propagation conditions. On November 3 stations will change over to their Winter/Summer schedules, so the information in this column will be liable to change on that date. Now onto this month's propagation predictions.

West Africa: 0800-1400 25, 21, 17 and $15 \mathrm{Mc} / \mathrm{s}$; 1400-1600 25, 21, 17, 15 and $11 \mathrm{Mc} / \mathrm{s} ; 1600-180025$, $21,17,15,11$ and $9 \mathrm{Mc} / \mathrm{s} ; 1800-200021,17,15,11,9,7$, 6 and $5 \mathrm{Mc} / \mathrm{s} ; 2000-020017,15,11,9,7,6,5$ and $4 \mathrm{Mc} / \mathrm{s}$; 0200-0600 15, 11, 9, 7, 6, 5, 4 and $3 \mathrm{Mc} / \mathrm{s} ; 0600-0800$ 15, 11, 9 and $7 \mathrm{Mc} / \mathrm{s}$.
South Africa: 0800-1400 25, 21 and $17 \mathrm{Mc} / \mathrm{s} ; 1400-$ $160025,21,17$ and $15 \mathrm{Mc} / \mathrm{s} ; 1600-180025,21,17,15$ and $11 \mathrm{Mc} / \mathrm{s} ; 1800-200021,17,15,11,9,7$ and $6 \mathrm{Mc} / \mathrm{s}$; 2000-0200 17, 15, 11, 9, 7, 6 and $5 \mathrm{Mc} / \mathrm{s} ; 0200-040015$, $11,9,7$ and $6 \mathrm{Mc} / \mathrm{s}$; $0400-0600 \mathrm{15}, 11,9$ and $7 \mathrm{Mc} / \mathrm{s}$; $0600-080021,17,15$ and $11 \mathrm{Mc} / \mathrm{s}$.
East Africa: 0800-1400 25, 21, 17 and $15 \mathrm{Mc} / \mathrm{s} ; 1400-$ $160025,21,17,15,11$ and $9 \mathrm{Mc} / \mathrm{s} ; 1600-180025,21$, $17,15,11,9$ and $7 \mathrm{Mc} / \mathrm{s} ; 1800-240017,15,11,9,7,6$ and $5 \mathrm{Mc} / \mathrm{s} ; 2400-020015,11,9,7,6$ and $5 \mathrm{Mc} / \mathrm{s} ; 0200-$ $040011,9,7$ and $6 \mathrm{Mc} / \mathrm{s}$; $0400-060011$ and $9 \mathrm{Mc} / \mathrm{s}$; 0600-0800 $25,21,17,15$ and $11 \mathrm{Mc} / \mathrm{s}$.

South Asia: $0800-120025,21,17$ and $15 \mathrm{Mc} / \mathrm{s}$; $1200-$ $140025,21,17,15$ and $11 \mathrm{Mc} / \mathrm{s} ; 1400-160021,17,15$, $11,9,7$ and $6 \mathrm{Mc} / \mathrm{s}$; $1600-180017,15,11,9,7,6,5$ and $4 \mathrm{Mc} / \mathrm{s} ; 1800-200015,11,9,7,6,5$ and $4 \mathrm{Mc} / \mathrm{s} ; 2000-$ $020011,9,7,6,5$ and $4 \mathrm{Mc} / \mathrm{s} ; 0200-0400 \mathrm{11}, 9,7$ and $6 \mathrm{Mc} / \mathrm{s} ; 0400-060011$ and $9 \mathrm{Mc} / \mathrm{s} ; 0600-080025,21,17$, 15 and $11 \mathrm{Mc} / \mathrm{s}$.

South East Asia: 0600-1000 25, 21 and $17 \mathrm{Mc} / \mathrm{s}$; 1000-1200 25, 21, 17 and $15 \mathrm{Mc} / \mathrm{s} ; 1200-140025,21$, 17,15 and $11 \mathrm{Mc} / \mathrm{s}$; 1400-1600 25, 21, 17, 15, 11, 9 and $7 \mathrm{Mc} / \mathrm{s} ; 1600-180017,15,11,9,7,6$ and $5 \mathrm{Mc} / \mathrm{s}$; $1800-$ 2000 15, 11, 9, 7, 6 and 5Mc/s; 2000-2200 11, 9, 7, 6 and $5 \mathrm{Mc} / \mathrm{s} ; 2200-240011,9$ and $7 \mathrm{Mc} / \mathrm{s} ; 2400-020011$ and $9 \mathrm{Mc} / \mathrm{s} ; 0200-050011 \mathrm{Mc} / \mathrm{s}$ only; $0500-060021,17$ and $15 \mathrm{Mc} / \mathrm{s}$.

North East Asia: 0800-1200 25, 21, 17 and $15 \mathrm{Mc} / \mathrm{s}$; 1200-1400 17, 15 and $11 \mathrm{Mc} / \mathrm{s} ; 1400-240011$ and $9 \mathrm{Mc} / \mathrm{s}$; $2400-060011 \mathrm{Mc} / \mathrm{s}$ only.

Australia via Asia: 0500-1000 25 and $21 \mathrm{Mc} / \mathrm{s} ; 1000$ 120021,17 and $15 \mathrm{Mc} / \mathrm{s} ; 1200-140021,17,15,11$ and $9 \mathrm{Mc} / \mathrm{s} ; 1400-180017,15,11,9,7$ and $6 \mathrm{Mc} / \mathrm{s} ; 1800-$ 2000 15, 11, 9,7 and $6 \mathrm{Mc} / \mathrm{s} ; 2000-220011$ and $9 \mathrm{Mc} / \mathrm{s}$; $2200-2400 \mathrm{11Mc} / \mathrm{s}$ only; $2400-0500$ Circuit closed.

## AFRICA

South Africa: R. R.S.A. Johannesburg is now using for its Afrikaans Service on Sundays from 0756-0950 25,790, 21,535, 15,220 and 11,900. English to New Zealand from 0956-1050 is on 21,535, 17,805 and 15,220. English to Africa from 1056-1150 is now on 25,790, 21,535, 15,220 and 11,900. English to Australia from 1156-1250 is now on $21,535,17,805$ and 15,220 . English Broadcast to Africa then run from 1256-1450 on $25,790,21,535,15,220$ and $11,900,1600-1650$ on 11,900 and $9,525,1700-1750$ on 21,535 and 17,805 , $1800-1850$ on 21,535 and 17,805 , to West Africa from $2100-2150$ on 21,535 and 17,805. The North American Service from $2326-0320$ is now on $15,220,11,875$ and 9,705.

ASIA
Dem. Rep. Vietnam: The Voice of Vietnam, Hanoi has now a Daily English Broadcast from 2000-2030 on $7,416,10,224$ and 15,018 , beamed to Asia, Africa and Europe.

## NORTH AMERICA

U.S.A.: According to information received by Mr. C. Pearson the $A F R T S$ is now on the following schedule:

New York Service from 1330-2300 on 15,430, 13301630; on $21,600,1630-2300$ on 21,605 -all beamed to Europe. A 24 -hour service is beamed to the Carribean area from 0600-1200 on 9,755 and 1200-0615 on 15,330. The West Coast Service from Los Angeles is now from $2130-0400$ on 21,$500 ; 2130-0730$ on 17,765; $0300-0800$ on 15,$410 ; 0600-1630$ on 11,805; 0800-1630 on 9,700 . Relay from Poro, Philippines to S.E. Asia 0645-1030 on 15,250.

## EUROPE

Sweden: Radio Sweden is now on the following schedule until Nov. 3, 0445-0615 on 17,840; 06300715 on 6,065 ; $0830-0900$ on 17,800 and 11,765 ; $0930-1030$ on 21,690 and 9,625 ; 1030-1100 on 9,625 ; $1100-1215$ on 15,315 and 9,$625 ; 1230-1330$ on 21,690 and 15,$420 ; 1400-1530$ on 21,585 and 17,$840 ; 1600-$ 1700 on 21,585 and 15,$310 ; 1730-1800$ on 15,240 and 6,$065 ; 1800-1830$ on 15,$240 ; 1830-1930$ on 15,240 and 11,$705 ; 1945-2015$ on 6,$065 ; 2015-2115$ on 11,915 and 6,$065 ; 2130-2230$ on 11,705 and 6,$065 ; 2245-2345$ on 11,705 and 11,$790 ; 2400-0230$ on 11,805 and 11,705 ; $0300-0430$ on 11,705. English is heard for 30 minutes at $0515,1100,1230,1400,1600,1900,2045,2245$, at 0030 and 0200 on 11,805 only, and at 0300 . There is also a Medium Wave English Service Daily at 22452315 on 1,178 , extended on Saturday night until 2430.

That's all for this month, deadline this month for all news is the 15 th, so good listening and 73 's.

GUESS what your Uncle David got this month. Two logs for forty metres and one for four metres-yes, $70 \mathrm{Mc} / \mathrm{s}$. So there is life in human form up there after all. Sad news for those who thought the world ended at $30 \mathrm{Mc} / \mathrm{s}$.

A fantastic pile of letters this month, so many that the local postman is claiming from me for his hernia! Let's start off with the comments and remarks of interest. The case of Jan Meyen Island and its prefix still rages. JX is Jan Meyen says K. Haywood (Lancs.) but what's NU3FYS? (The wittiest remark received will be rewarded with a dud OC71). Yes, JX is the Jan Meyen prefix warbles S. Pitt (Essex). Definitely, reckons C. Morgan (Northumberland). No, JX is the Svalbard Archipeligo JW is Jan Meyen says R. Sherwin. (Oh Gawd, I wish l'd never mentioned the business). C. Morgan queries LG5LG claiming to be-no never mind let's sort Jan Meyen out first.

Looking round the six bands generally they all seem in pretty good shape except 10 metres which stays pretty quiet at all QTH's. Fred, GC3SVK/P, was logged on topband by quite a number of listeners but some said that he was still in Guernsey and not Jersey on.the 7th. Shame on you Fred, what was her name?

Twenty metres offered some very good DX and remained open for quite long consistent periods. Oceana came through most mornings around 0600 while South America peaked up in the late evenings about 2200 . Fifteen appeared very good at times but patchy. Certainly there was no shortage of African and S . Americans.

## FOUR \& FORTY

"Thanks for your mention of the four metre contest" says J. Ashworth (Lancs.). He has a converted R1132A with crossed dipoles (NS and EW) at 30ft. He logged 126 stations which included EI2W (20W to a 4ele), G3JIJ/P ( 100 mW from Ingleborough Hill Summit, Yorks.), G3RLE/P ( 100 mW from Pen-y-ghent summit, Yorks.), GW3ERB/P (15W, Halkyn Mountain, 950 ft . a.s.1.), GW3ITZ/P ( 5 miles S.E. of Flint), GW3NUE/P (Pen-y-gaderfawl, 2,600ft. a.s.l.), GW3OXD/P(Radnor). John says that the main spots of four metre activity seem to be Southport, Liverpool, and Wirral.
F. McVerry (Lanarkshire), R1155, 75 ft . long wire at 20 ft . running NS sends in a FB log for $7 \mathrm{Mc} / \mathrm{s}$ - CM 2 DC , CO3NR, CN8AW, CR6JW, CT2AP, EA6BG, EA8FF, FØKC, GB3SH, GC3GS, HBØAAI, ITIKAT, JX2BH (no comment), K7WWR, KL7FLG/P/3, LJ2X, LU2DGO, LXISL, OK8AAA, OX3DX, OY2H, OY5NS, PIINTB, PY6HL, PZICF, TA2RX, TF3OM, UL7QO, VE1AMB, VE3OE, VK7SM, WIBWN, W5OVR, W9AQW, YVIAP, ZC4RB, ZD9BE, ZSIJA, ZP3AB, $3 \mathrm{~V} 8 \mathrm{AA}, 4 \mathrm{X} 4 \mathrm{WN}, 5 \mathrm{~N} 2 \mathrm{AAX}, 6 \mathrm{~W} 8 \mathrm{AL}, 9 \mathrm{Y} 4 \mathrm{KR}$.

Francis tells of G3SWZ working some 30 countries on $7 \mathrm{Mc} / \mathrm{s}$ with his $\frac{1}{2}$ watt rig. Vy FB indeed.

Martin Pasek (Notts.) QP166 into an HRO, "40 metres $X$ beam made of wire" (Cor)-LU9DM, PZ1AX, SMOKN/Ø, VK2ABZ, VK2AYC/M, VK3AC, W2RFS, YU2AKL, ZL2BCG. Martin says that ZL2BCG is also on eighty s.s.b. around 0600 and then QSY's to $7.097 \pm 3 \mathrm{kc} / \mathrm{s}$ at 0615 . The VK's peak around 0730.

## FIFTEEN

J. Baker (Lancs.) PCR3, 60ft. long wire at 18 ft . running NW/SE-CE3FI, CN8BV, CO8RA, CP6EL, CR4BH, CR6LL, CR7IC, CT2AC, EA6BC, EA8FF, ET3REL, FM7WE, HC2LM, HI3AGS, HK1NR, HP4BIW, HR1KS, JAIIJH, JA2IIK, JA6DZA, KG4AM, KP4AOW, KV4FC, LU2AHI, OA5AM, OD5EP, PY2DPG, PY7OS, PZ1BX, SVICD, TJ1AL, TU2BQ, VE2BAG, YV3FB, ZC4MO, ZD8Z, ZE2KL, ZSIBV, ZS5BK, ZS6IN, 3V8AA, 4X4ZL, 4Z4HG, 5A4TH, 5N2AAF, 5 W 1 AS , 5 Z4LG, 6 W 8 AL , 7 Q 7 BM , 8P6BC, 9 GIFL , $9 \mathrm{GIGD}, 9 \mathrm{H} 1 \mathrm{~K}$, 9 K 2 CC , 9 M 2 PO , 9Q5LC, 9Y4CR, 9X5LR.
P. Starling (Essex) HRO, 30ft. of wire at 15 ft .CR6DP, EL4C, EP3MAM, ET3REL, JT1GF, KøBSZ, K1GUD, OD5BZ, PY5EC, ZC4GM, 4X4BL, 4Z4HO, $5 \mathrm{~A} 1 \mathrm{TG}, 5 \mathrm{~A} 3 \mathrm{TX}, 5 \mathrm{~A} 4 \mathrm{TH}, 5 \mathrm{Z} 4 \mathrm{LG},{ }^{2} 9 \mathrm{H} 1 \mathrm{BG}$, 9 K 2 CB , 9K2CC.

## TWENTY

C. Morgan (Northumberland), CR150/2, 25 ft . vertical, s.s.b.-AP2MR, DU1DBT, HC4PS, HP1LV, JH1DHQ/MM, K5ZSX/AM, K8LSB/P/KG6, KA2NA KH6BX, KL7EZM, KR6EB, KV4DB, PJ3CC, PYIBOY, PZIAC, VK2ALR, VK5XV, VK6NM, VK7DK, VK9XI (Christmas Island), VP7NP, VP8HZ, VQ8CG, VR6TC, VS6DR, VU2DKZ, W4UAF/P/ KH6, W7DLF, WA2FWQ/MM, WA6GZZ/AM, XE1KV, YN1GLB, ZL3UY, ZS5DC, 4Z4HF, 5H3KJ, 8P6AU, 9M2BK, 9NIMM, 9VINR.
H. Norman (Suffolk), modified Vidor CN385, 45 ft . end fed with an a.t.u. (sensible chap)-CE3BE, OA4TN, OX4WN, PJ9AQ, PY3GV, SK4AZ, VEIASJ, VE3NL VE6AQL, VKIRB, VK2SU, VK3VK, VQ8BCI, XE8UBL, YV5CNQ, ZP3BW, 4X4DL, 5A2EF, 5Z4KL, 9G1GD.
G. Kelly (Liverpool), R107, 35ft. vertical-all s.s.b.CE1IU, CPIAP, CN8MT, CR6CA, CR6CA, CR7CH, CT2NAA, CX9AN, DU1HR, EL2AS, HC4WN, HC5MP, HI3ELJ, HPIQH, HR6EB/P/2, HS3MJ, JA1AE, JX1OM (snigger), JX3DH9 (titter), K1FNA/ P/KG6, KL7AHB, KR6NR, LU2CF, LU8PM, MP4TCE, OA4BS, OX5AY, PXIYY, PYITX. SVØWGG, TI2LJ, TU2BQ, VE8RCS (Weather Station on Ellesmere Is.), VE8ML, WB4IRT/AM, YEIDAN, YV4IIQ, ZB2AY, ZD8CC, ZP5CF, 9Y4VT, 806AH, 7X2ED.
R. Street (Surrey) 6 -valve s'het, 5 RV at 10 ft .CO2FA, CP4DA, EL2Z, FO8BY, HS3ZZ, JXIOM, K7DSB/P/KH6, VP2IA, VS6DR, YAIAP, YK1AM, ZP5JB, 4A2IH, 5U7AL, 8P6BX.
I. Stevenson (no address), GC-1U, 50 ft . end fedCTIMW, FЯMI/FC/M, HBØLL, HV3SJ, K6UJW, PY7AOA, PXIUB, VE2WY, VK2SG, VK3ABE, VK4KS, WA6EWI, WN6OIV, W6CCB, W7LFA, YV5BDA, ZL3QM, ZL4BX.

## CONTESTS

There's a few contests about in October; 5-6,70cms contest; 12-13, ten metre phone contest, how about listening? 12-13, 1,296Mc/s contest; 19-20, Jamboree on the air: 26-27, 40 metre c.w. contest. Deadline for logs is 14 th of this month, good listening.


THE tuner is an exceedingly simple unit, giving medium and long wave coverage. With such a tuner, results depend largely on local conditions, distances to stations, and the aerial. Usually, two or more BBC stations can be expected with an indoor aerial and earth, or an outdoor aerial without earth. Volume is greatly increased by employing a fairly good outdoor aerial and earth.
If the tuner is used with the "Pyramid" amplifier, very good volume can be obtained. Such a tuner is not intended for long distance reception, or in circumstances where the single tuned circuit does not give enough selectivity. The tuner may be used with other amplifiers, or as a crystal radio, operating headphones. A reasonably effective aerial and earth will then be needed.

## Construction

Figure 1 is the circuit of the tuner. The tuned windings are on a $4 \times \frac{3}{8}$ in. diameter ferrite rod, as this allows an efficient coil to be made with minimum difficulty. The medium wave section, from A to C, has 55 turns of 28 s.w.g. double cotton covered wire, about $\frac{3}{8} \mathrm{in}$. from one end of the rod, and wound in a pile to occupy about $\frac{1}{2}$ in. The aerial tapping B is 15 turns from C. The long wave section, between points $D$ and $E$, has 140 turns of 32 s.w.g. silk covered wire, and occupies a pile about $\frac{3}{8} \mathrm{in}$. long, 2 in . away from the medium wave section. The windings are connected at C and D so that all turns throughout (that is, from A to E) are in the same direction. Other wire gauges could be used, or a ready-made coil for medium waves, or medium and long waves might be selected.
The parts are assembled in a plastic box about $4 \frac{1}{2} \times 4 \frac{1}{2}$ in., Fig. 2 . Holes should be drilled with care, as the material may be brittle. For this reason,
large holes for switch and variable capacitor are best made by drilling a ring of small holes.
A co-axial lead takes output to the amplifier, the outer part of the plug and outer screening forming the earth return circuit. For headphone reception as a crystal set, two further terminals could be fitted instead -one connected to earth, and the other to the diode. Medium or high impedance phones ( 500 to $4,000 \Omega$ ) are most suitable.

## Aerial

Though used remote from any BBC station, it was found that a 15 ft . indoor aerial gave enough volume, when an earth was added. An insulated wire can be stretched along one or two walls of a room, near the ceiling. Extremely short or poor aerials can be connected directly to point A.

Outdoor aerials are erected in the usual way, and


Fig 1: Circuit of the crystal diode turier.
may be some 20 to 50 ft . or so of $7 / 26$ or other aerial wire, according to circumstances. An outdoor aerial may be required if the tuner is used with an amplifier having less high gain than the one described.

Medium wave coverage is obtained with the switch closed or "on" and long waves with it open or "off". With long aerials, tuning may be made more selective by including a small capacitor in the lead to the aerial terminal.


Fig. 2: Construction and wiring details of the crystal diode tuner.

## components list


#### Abstract

DIODE TUNER 500pF variable capacitor; knob; on/off switch; OA81 or other diode; $4 \times \frac{3}{6}$ in. ferrite rod; 28 s.w.g. and 32 s.w.g. wire; two terminals; plastic box; screened or co-axial lead and plug.


## Transistor Morse Oscillator

This is a single transistor oscillator producing an audio tone for Morse code practice. It is intended to plug directly into the "Pyramid" amplifier for loudspeaker output. It could be used with other amplifiers, or by itself with headphones.

Figure 3 is the circuit, and a small driver type receiver transformer is employed for feedback, one connection of the tapped secondary being unused. Should a different transformer be fitted, connections to one winding may have to be reversed, to obtain oscillation.

Since the amplifier mentioned has high gain, output is taken from the oscillator emitter, Fig. 3. For headphones, or amplifiers requiring more input, the audio signal can be taken from the collector, through a $0.1 \mu \mathrm{~F}$ or similar capacitor, as in Fig. 3. The amplifier is returned to the positive line of the oscillator with the equipment described, this connection is made by the outer part of the co-axial plug.

## Construction

The parts are assembled on a small paxolin panel, Fig. 4. Dimensions could be changed to suit a small plastic box or case to hand. Components occupy one side of the panel, and wiring is carried out on the reverse. Two flexible leads run from the points shown, and are taken to the key terminals. Connections to the amplifier can be by co-axial screened lead, or twin flex. The oscillator emitter goes to the amplifier audio gain control, the co-axial inner conductor being used for this purpose.
As a 1.5 V supply was found most suitable, and current required is very small, a single dry cell is soldered directly in place. The zinc case is negative. No switch is present because current is only drawn when the key is closed. All transistors tried worked satisfactorily, including cheap surplus audio types.

## Audio Tone

It is only necessary to plug in as described, and adjust volume with the amplifier control. Adequate volume, according to circumstances, will result in an amplifier battery drain of about 15 to 25 mA (key down). It is as well not to have volume so great that current exceeds 50 to 60 mA .
If a different transformer is fitted, and the audio tone needs to be changed, the effect of using a different battery voltage can be tried. The transistor or resistor values can be changed, or a capacitor placed across one transformer winding, to reduce the frequency. These would be a matter for experiment.


Fig. 3: Morse oscillator circuit.

## Notes on Use

The audio oscillator can be used to monitor outgoing c.w. transmissions by deriving its operating current from a diode and r.f. pick-up loop. It may provide a signal in phones, or may be used with the amplifier for speaker reproduction. Either method allows a check on keying.

The "key" leads are joined together, and the battery removed. Current is then obtained from the
loop and diode in Fig. 5, which is connected in place of the cell. A pick-up coil having five or six turns, 1 to 2 in . in diameter, should prove satisfactory. This can be modified, if necessary, to suit the transmitter


Fig. 4: Oscillator wiring and layout.


Fig. 5 (left): Loop and diode for c.w. monitoring.
power, etc. The amplifier should be switched on, and the loop approached to the transmitter tank coil or aerial tuner, until a suitable audio tone results when the transmitter is keyed. Avoid taking the loop too near the source of r.f. energy.

Not only may this damage the diode DI, but direct contact with the tank coil could be disastrous.

## * components list

## MORSE OSCILLATOR

Paxolin $2 \times 2 \frac{3}{4} \mathrm{in}$; audio transistor; $\frac{1}{4}$ watt resistors$1 \mathrm{k} \Omega, 4.7 \mathrm{k} \Omega, 27 \mathrm{k} \Omega ; 0.04 \mu \mathrm{~F}$ capacitor; QXD1 transformer (Osmor); $1 \frac{1}{2} \mathrm{~V}$ dry cell; screened or co-axial lead with plug.

## TO BE CONTINUED

## CORRIGENDA <br> PYRAMID-ALL PURPOSE SYSTEM <br> SEPTEMBER ISSUE

In Figure 1, Tr 4 was drawn incorrectly. The emitter should go to R14 and the collector to the junction of R15/T2. R1 in the components list should read $2.7 \mathrm{k} \Omega$.


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# practically Wireless commentary by ILINII 

HAVING trouble with your new decimal coinage? Wondering how may sides make seven? You don't know your luck, lads. To radio enthusiasts, the change from archaic, rule-of-thumb measurement to our modern, swinging, polished, simply-divisible, etc., etc., should come easy. We have been dealing in multiples and divisions of ten for longer than de Gaulle has been a General.


It was not so long ago that I learned about the origins of Arabic numerals-those squiggly symbols on your Bingo card. Ever ready to widen the cultural interests of my readers I pass on the information that the shapes of the originals dreamed up by a great Algerian mathematician each contained the number of angles relevant to the numeral-if you drew it properly. We have now stylised them out of any semblance to the first idea but working them out can be quite fun.

Have you ever thought of the cogitation that must have gone into our component colour code? Imagine a committee sitting around a polished table, perspiration laving their bald pates, battling over the redness of two or the royal blue impeccability of six. Come to think of
it they did not do a bad job. I mean black is indisputably "zero" and "one" has a brown dinginess about it, the magical number seven has an emperor's aura just as three must be orange; couldn't be anything else.

Our active P.A.L. protagonist Bernard Rogers of the Rank-Bush-Murphy research establishment has described the multilingual vote-chopping that went on while colour-TV systems were being chosen. According to him we in Britain are lucky to have settled on a system at all. Perhaps it is as well that the colourcode was already settled before colour-scales had to be agreed. (Although to be honest one wonders whether there is any real standard when one sees sets of rival manufacture receiving the same programme side by side-in a colour-TV workshop where setting-up has been done with meticulous care.

The chap in his parlour with no comparisons to make except the clash of Lulu's rosiness with the missus' choice in upholstery may not hit on this hue-change problem very quickly. But when the novelty has worn off look out for: "Let reds down a bitwhoa! now a touch of yellow in the greens," etc. Every television engineer will have to carry his illuminated colour-scale in the lid of his toolbox.

Of course the drawback is in those little qualifications "to international standards". There are no such animals. When in Rome the yellows and magentas look a good deal different from Aunt Harriet's puce and poppy at Scarborough. Blue skies over the Côte d'Azur would seem too dramatic for Streatham.

And in any case it is doubtful whether international standards would be gladly accepted by all. Just consider the recent hullabaloo at the Royal Society, when the Conference of Editors
called for the adoption of SI in all scientific and technical journals. SI-Système International d'Unites-is a system of refinement of the original metric figuring officially adopted in 1960 and the legally accepted system in thirty countries. Britain is still testing the water-she has so much to lose. All those lovely avoirdupois units, and the Imperial measurements being


Stand on our heads again
consigned to a museum. It is unthinkable. We should have done much more lobbying. After all, the astronomers managed to retain the parsec ( $30.87 \times 10^{15}$ metres) and that's not a measurement you are likely to come across every day. More important to us, the electrical engineers lobbied successfully for the electronvolt, which is $1.6021 \times$ $10^{-19}$, so you needn't forget that.
We radio bods have been talking internationally for longer than we care to remember. Change comes natural to us, and not only in the metric scale. Look at the sudden flip to all our ideas that transistorisation brought about. Negative lines at the top, forsooth! And now the F.E.T. boom is swinging along, and we shall have to stand on our heads again. Never mind, Joe, just so long as you remember a pint is 0.0005682615 of a cubic metre


INTHE
HOIME
MARTIN L. MICHAELIS, M.A.

IN the course of time, the keen audiophile will accumulate many pieces of equipment which may be home made, commercial products or assembled from kits. These all tend to get tagged-on to the existing rig in the family living room in some haphazard manner. One day the wife will find that the rules of interior decoration have been abused beyond all tolerable limits. This article describes in quite general terms, the action taken by the author when this justified protest was voiced. High-fidelity sound reproduction equipment is no longer a novelty fascinating the owner and visitors alike, regardless of appearance as long as it sounds good. It has now become a commonplace feature of home life and styling is as important as the acoustic performance. Furthermore, the operating procedure for a composite equipment must be so simple that any nontechnical member of the family can switch on his favourite entertainment. The necessary mains switches for the many items of equipment pose a special problem. These must be arranged such that it is not possible to leave any piece of ancillary equipment switched on indefinitely without this becoming immediately obvious.

We are not setting out to give new hints for the connoisseur to enable him to obtain just that extra response which no average person is able to hear anyway. We are addressing the average person, who will demand a good standard of reproduction, ample but not pedantic response at both ends of the frequency range, and versatile entertainment with a domesticated equipment. This adjective refers to home entertainment as the prime aim, as distinct from perpetual electronic patchwork. Experimenting is envisaged once and for all whilst installing the domesticated high-fidelity system, for which this article gives a few collected hints.

## Functions

A complete installation will generally be called upon to cater for television, sound broadcast and recorded sound entertainment. The latter must cover tape and gramophone records, with facilities for transcription from all other sources on to tape. All functions except television should be capable of stereophonic operation. Finally, it should be possible to relay all sound programmes, whatever their origin, to one or more extension speakers in other rooms.

## Equipment

The basic equipment which will provide these functions comprises a television receiver and a modern radio tuner unit. The latter will cover all
desired a.m. wavebands and the v.h.f.-f.m. waveband incorporating a stereo decoder. A stereo amplifier can be built-in or provided as a separate unit.
A number of ancillary units is necessary. The internal amplifier and speaker of the television receiver are often inadequate for optimum reproduction of the television sound, so that some form of auxiliary amplifier and linkage to the main system are desirable. The radio tuner unit will be operated in conjunction with a stereo tape recorder and record player, and a stereo amplifier if not already incorporated in the tuner unit.

## Loudspeakers

For all kinds of sound reproduction in the living room, it is desirable to use one and the same pair of high quality loudspeaker box units, in the usual stereophonic arrangement. Monaural radio, tape or gramophone record programmes should be reproduced with equal intensity on both loudspeakers. This enormously enhances the presence of any monaural programme, compared to single-speaker reproduction thereof. It is important to use the same separation between the two loudspeaker boxes, and to sit at the same position as for good stereophonic listening, if maximum benefits are to be exploited for monaural listening too.

Television sound is an exception to this rule, because it is necessary to ensure subjective coincidence between the picture and the sound source. The only way to achieve this simultaneously for all viewing positions, is to stand the television receiver on one of the loudspeaker boxes, and to arrange for the TV sound to be boosted only through this loudspeaker.

Two major questions concern the extension speakers of the composite system. Firstly, the programme running in the living room may be monaural or stereophonic. In either case, the extension speakers must bring a properly combined monaural version which is the sum signal of the two stereophonic channels in the living room, without disturbing the stereophonic reproduction there by introducing crosstalk between the channels or other undesirable effects. The extension speakers will be situated in rooms such as the kitchen or the -nursery, where stereophonic reproduction would hardly serve any useful purpose.
Secondly, much smaller speakers serve for the extension positions. These can certainly produce the same subjective volume as a large box, but only for a restricted frequency band. Above all, they are unable to handle bass frequencies at the same relative levels as required for a large box unit. If the frequency range is not restricted, the extension speakers will hoot and rattle due to frequency selec-

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tive amplifier distortion and excessive speech cone excursions, when the signal is correct for the large box units in the living room. On the other hand, if the common tone controls are adjusted to give pleasant reproduction at the extension speakers, the quality at the large box units would be unacceptably thin.

## Separate amplifier

It is hardly possible to meet these basic requirements for the extension speakers without providing a separate amplifier for the purpose. This is also expedient in relation to the output loading characteristics of many types of transformerless output stages in modern transistorised amplifiers. These stages operate at high power efficiency and low distortion, but require critical load matching. The output power is drastically reduced if the impedance of the connected speaker is significantly greater than the optimum value. On the other hand, if the loudspeaker impedance is less than the optimum value, the power transistors are soon destroyed on loud
passages. If provision were to be made for arbitrarily switching extension speakers in parallel with the main speakers, as has been customary with valve output stages, it would thus be necessary to make the impedance of the main speakers considerably greater than optimum, so that the net load impedance presented to the amplifier just drops to the optimum when all extension speakers are switched on. This means that the main loudspeakers can never receive the full peak power. It is not primarily volume, but transient response which then suffers. Difficult passages of complex polyphony, or powerful instruments such as an organ, will sound confused and unrealistic.
It is convenient to derive the input signal for the separate amplifier feeding the extension speakers, from the output of the main amplifier across the main speakers. The auxiliary amplifier will incorporate only preset controls, to achieve correct tone and volume matching between the main speakers and the extension speakers, and to effect re-combination of a stereophonic signal to a monaural sum signal for the extension speakers. The externally accessible


Fig. 1: Suggested interconnections of Radio, TV Receiver and Ancillaries. The auxiliary TV sound amplifier, the auxiliary amplifier for the extension speakers, the relay circuits and the associated hum-loop isolating transformers and resistors are all built into the cabinet of the right channel loudspeaker, which also carries the control knob of the extension speaker selector switch S1 on its front. The following input cable's run to this composite unit: switched mains from radio tuner; switched mains from TV receiver; input mains line; combined loudspeaker output cable of radio tuner/amplifier; television sound output from TV receiver. The following output sockets are fitted to the composite unit: feed to left channel loudspeaker box; feed to extension speaker A; feed to extension speaker B; TV sound for tape recorder (monaural); radiolphono sound to tape recorder (stereo). Thirdly, the following mains out/et sockets are fitted to the composite unit: for radio tuner/amplifier; for television receiver; for tape recorder; for record player. The tape playback cable (not shown) runs direct from the tape recorder to the radio tuner/amplifier in the usual manner.
controls of the main amplifier (radio tuner) then serve jointly for all loudspeakers in the entire system.

## Switching and manual controls

A particularly confusing feature of arbitrarily assembled systems is the multiplicity of uncoordinated controls on the numerous separate units. These have to be set in certain mutual ways to give correct performance, other combinations of settings making the equipment appear quite dead or even causing damage to it. An example of the latter danger is an arbitrary switching system connecting various loudspeakers to the output of one of the aforementioned load-sensitive amplifiers. The experimenter will take care to avoid forbidden switch settings, but he can hardly expect this of other members of his family, who have an equal right to use the equipment in his absence -and a right to expect it to work immediately after comprehensible actuation of few controls, without going up in smoke. Forbidden mutual settings which could cause damage to the equipment should be made impossible by using internal auxiliary relays and restricting the number of externally accessible controls.
Another confusing form of manual controls is the use of more than one control for the same function in any single channel. For example, a record player unit may be fitted with a volume control. A separate equaliser preamplifier unit will possess another volume control. The main amplifier may possess a third volume control. If all three volume controls are externally accessible, somebody need only have inadvertently turned any of them to zero, whereupon the other two will fail to produce a signal and the owner, upon coming home, is greeted with "your rig has broken down again". In a well-designed installation, each type of entertainment (radio, television, tape, records) should be governed by only one manual control for each function, and it is desirable to share as many controls as possible between several branches of entertainment.

The third common source of confusion is the mains on/off switching. The non-technical user of the equipment here demands an intuitively obvious switching arrangement. He correctly classes sound entertainment into two broad groups: radio and television. If the form of entertainment he desires is connected with television, he wishes to actuate a switch on the television receiver, nothing more. Auxiliary amplifier switching, loudspeaker selection and any necessary muting of other sources should be effected automatically with relays in the concealed wiring between the units. If the desired form of entertainment is connected with radio, including tape and records, switch-on should be actuating a mains switch of the radio tuner unit. The record player mains switching will be controlled by the pickup arm, and the tape recorder will have its own mains switch. The mains feed for both units should be controlled by the mains switch of the radio tuner. This ensures that everything is definitely switched off when both the radio tuner and the television receiver are switched off. These two principal units usually possess some form of pilot lamp or dial lights, so that it is obvious at a glance whether or not they are switched on. Modern transistorised radio tuners are usually compact and light, so that one very suit-
able position for this unit is standing on top of the television receiver. It may be advisable to make a small modification to the tape recorder to ensure that the drive mechanism is automatically disengaged when mains power is switched off at some external point. This prevents the appearance of an eccentric position on the rubber pressure wheel of the drive capstan. The simplest arrangement is to use a resting contact of a mains energised relay to discharge a capacitor through the same actuator solenoid as used to disengage the mechanism automatically at the end of the tape. A working contact of the same mains relay holds the capacitor charged as long as the mains supply is present.

## Fuses

Commercial radio tuners and television receivers do not normally provide a mains outlet via the internal mains switch, so that this simple modification will have to be added. Make sure that the switch contacts are of adequate rating; otherwise fit an auxiliary relay. Do not run the switched output line via the existing internal fuse, but add a second fuse of appropriate rating in the output branch line beyond the switch. A common fuse for the internal circuitry and the new equipment connected externally would have to be up-rated to carry the combined load, which would seriously impair the degree of individual protection given for each item.

## Switching the extension speakers

We have already seen that the extension speakers require their own audio amplifier. This should be switched on only when at least one extension speaker is actually in use. The author has found the following arrangement to be neat and convenient, as well as never giving rise to confusion of non-technical users.

The auxiliary amplifier is fitted with two relays and a three-pole four-way switch. The relays possess mains voltage coils rated for continuous duty. One relay coil each is connected straight to the switched mains output of the radio tuner and the television receiver. The relays each possess one maker contact, these contacts being wired in parallel and mutually in series with the mains switching wafer of the four-way switch. The mains feed is taken via this combination to the mains transformer of the auxiliary amplifier. The other two wafers of the four-way switch connect the extension speakers for two rooms (kitchen and nursery). Three positions give the functions kitchen/nursery/both and the fourth is the off position. The auxiliary amplifier obtains mains input only when the extension speaker switch is set to one of its action positions and either the radio tuner or the television receiver is switched on.

## Simultaneous operation

Simultaneous use of the television receiver and radio tuner must be considered. Firstly, it may be desirable to relay a radio, tape or record programme to the extension speakers whilst other members of the family are watching television. The converse is hardly sensible, so that it suffices to provide another

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| 1,600 | 2,500 |  |
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circuit energised from the switched mains outputs of the radio tuner and television receiver, sensing when both are switched on. In the latter case, the audio input from the television receiver to the auxiliy amplifier is muted by relay contacts. Only the radio output then reaches the auxiliary amplifier and is reproduced by the extension speakers.

Secondly, it may be desired to tape an interesting radio programme which is broadcast at the same time as an important television programme. This function is immediately possible when the radio tuner is switched on, irrespective of whether the television receiver is also switched on. A further logical distinction is generally also required. The tape recorder may be required to record the television sound. Thus its mains feed must be taken via the pair of relays on the auxiliary amplifier, so that it is available whenever either the radio tuner or the television receiver, or both, are switched on. Furthermore, the TV sound muting contact at the input of the auxiliary amplifier must also mute the television sound feed to the tape recorder when the radio tuner and the television receiver are both switched on.

With this arrangement, the tape recorder is ready to record television sound whenever the television receiver is switched on alone. If the radio tuner is switched on alone, or both units together, the tape recorder is always ready to record the output of the radio tuner, including the record player connected to it. This also means that taped programmes can be played back to the extension speakers whilst viewing television in the living room. If none of the tape recorder functions are required, the tape recorder is left switched off at its own mains switch. However, if this is forgotten, the tape recorder can never inadvertently be left switched on for days on end, because its mains feed is automatically broken as soon as both the television receiver and the radio tuner are switched off.

A monitoring facility is desirable in the living room for the second programme when simultaneous operation of the radio and television functions is
being used. This permits brief listening-in to the second programme being tape recorded or played out to the extension speakers, for making checks and adjustments.

For this purpose, it is convenient to exploit the existing pair of stereophonic loudspeaker boxes in the living room. The switched mains output of the television receiver is used to feed the mains solenoid of a further relay fitted with two-pole changeover contacts. At the same time, the switched mains output of the television receiver feeds a second auxiliary amplifier directly. The changeover contacts of the energised relay disconnect one stereo channel loudspeaker from the main radio tuner amplifier and connect it to the output of the television sound auxiliary amplifier. The television receiver is standing on this loudspeaker box. The other stereo channel loudspeaker is left connected and will reproduce the radio and tape programmes even when the television receiver is running. This other channel loudspeaker can be silenced by turning the balance control to the appropriate stop on the radio tuner. In this arrangement, the existing balance control of the radio tuner serves as monitor volume control for the second programme. It will not affect the outgoing volume to the extension speakers, because these are fed with a sum signal which is unaffected by the stereo balancing setting. The balance control always makes equal respective boosts and cuts in the two channels, leaving the sum signal almost unchanged.

## Block diagram

Figure 1 shows a block diagram of the complete equipment assembled according to the details discussed above. This is intended to be valid in general and thus emphasises the interconnections required to obtain the described intuitively obvious operating procedure. The choice of individual units (television receiver, radio tuner, amplifiers, etc.) is not subject to any fundamental restriction. Any design


Fig. 2: The inputs mixer section of the extension speaker's auxiliary amplifier. The section from T2, T3 to Tr2, Tr3 develops a properly summed monaural signal across R13, from the stereo input. C2, C3 in conjunction with VR2, VR3 provide the necessary bass cut (see text), whilst VR2, VR3 serve to balance the addition function and the volume at the extension speakers with respect to that at the main stereo boxes. The section T1 to Tr1 mixes-in the TV sound, the output again appearing across R13. The stages beyond the components R13, C7 are conventional and may use any convenient circuit to develop 4 watts output power. The input transformers are necessary to avoid hum loops. They may be miniature components:
published in other articles in this magazine, commercial items or kit constructions are suitable in principle. In this article we do not intend to go into basic circuit details of high-fidelity and stereophonic equipment, but some remarks concerning the most desirable features for a composite installation are appropriate. The most important question is the audio power of the respective amplifiers.

## Audio power ratings

The actually required input power to a loudspeaker for a given subjective volume is considerably influenced by the efficiency of the loudspeaker, by the size of the room and by the acoustic damp-


Internal view of the author's right-hand channel loudspeaker cabinet.
ing factor (furnishings). Modern high-fidelity loudspeakers often show a lower electrical sensitivity, which is the price paid for low distortion and brilliant transient response. However, the frequency is the dominant factor determining the required electrical power for a desired acoustic volume, and to a first approximation it is the only variable in this sense. Let us first of all consider the reproduction of pure sinewave tones at various frequencies. For the speech band or middle register, extending from about $200 \mathrm{c} / \mathrm{s}$ to $2 \mathrm{kc} / \mathrm{s}$, we find that an electric power of 2 watts is ample with an average loudspeaker in an average room. This is sufficient for very loud undistorted reproduction with an intensity of about 85 phon. Somewhat less than 1 watt suffices for loud reproduction, 250 mW for normal reproduction and about 50 mW for quiet reproduction, under the same conditions and for the same restricted frequency range.

Thus it is evident that the correct rating for the
extension speakers auxiliary amplifier is 4 W when two rooms have to be served, so that each speaker can receive 2 W .
The required electrical power increases greatly when it is desired to reproduce lower or higher frequencies with the same subjective volume. This is necessary for true high-fidelity reproduction in the living room. Whereas about 2 W were sufficient to produce 85 phon sound intensity at $1 \mathrm{kc} / \mathrm{s}$, about 14 W are necessary for this purpose at $30 \mathrm{c} / \mathrm{s}$, and about 10 W at $15 \mathrm{kc} / \mathrm{s}$. These powers may be halved for a stereophonic installation, as far as each channel is concerned, since both channels contribute jointly. In simpler equipments it is generally unnecessary to make provision for full peak volume from each channel alone, since the majority of stereophonic effects are produced by phasing, not amplitude differences. Both channels then contribute to the volume, even though the subjective sound source may appear to lie to one side. Taking these considerations into account, it is found that a main stereo amplifier with 8 W output per channel is fully adequate for an economical installation and gives ample performance for the average listener in the average living room. The critical connoisseur will not be satisfied. For him, a rating of at least 25 W per channel is necessary in a more sophisticated equipment. This power does not serve to give more volume; the 25 W amplifier will not sound much louder than an 8 W one. The difference becomes apparent with complicated polyphonic passages with their intricate transients. Since the average power never runs the 25 W amplifier close to the drive limit, smaller transient intermodulation effects are encountered and difficult instruments such as an organ will sound more realistic to the trained ear. The average musically untrained person will usually be unable to hear the difference, so that a twin channel 25 W stereophonic amplifier would involve him in unnecessary expense, a twin 8 W amplifier serving just as well at lower cost.

## TV sound amplifier

This leaves the second auxiliary amplifier for the television sound. The television receiver will already incorporate an amplifier loudspeaker delivering and handling some 2 to 4 W of audio power, but often lacking full bass. An appropriate rating for the auxiliary amplifier is therefore about 6 W , and it should incorporate preset controls to give little output above $1 \mathrm{kc} / \mathrm{s}$, so that the 6 W boost power are used predominantly to boost the bass and lower middle registers from $30 \mathrm{c} / \mathrm{s}$ to $500 \mathrm{c} / \mathrm{s}$. The internal amplifier and speaker of the television receiver are left running and the feed for the auxiliary bass-boost amplifier is taken across the internal loudspeaker. It has been found that this arrangement gives extremely good presence and coincidence of the sound in the picture, regardless of the viewing position in the room. It is preferable if the internal loudspeaker of the television receiver radiates through the side of the cabinet with reflection from one corner wall of the room, to fan out the treble. The bass comes from the high-fidelity box unit on which the television receiver is standing. A large bass reflex cabinet is appropriate and it may be used to accommodate the auxiliary amplifiers and relay circuits on internal shelves.

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## Frequency range matching

We have seen that about 2 W suffice in the middle register to produce the loudest required volume in any average loudspeaker. This corresponds to an r.m.s. drive voltage of about $2 \cdot 8 \mathrm{~V}$ for a $4 \Omega$ speaker. When the full frequency range boxes in the living room are running at maximum volume, middle register frequencies will develop this voltage across the speech coils (assuming $4 \Omega$ impedance). But extreme bass frequencies will be feeding some 14 W to the speakers, corresponding to nearly 8 V r.m.s. across the speech coils. This will immediately overload the auxiliary amplifier used for the -extension speakers, because it must be adjusted such that 2.8 V input from the speech coils of the boxes reproduces this same voltage across the speech coils of the extension speakers. The middle register, which dominates in determining the overall subjective volume, is then reproduced with the same intensity at the main loudspeaker boxes and at the extension speakers in the other rooms. The auxiliary amplifier is not designed to handle 8 V r.m.s. input at any frequency, since it cannot develop the corresponding output power. Thus the bass components must be suppressed before they enter the amplifier. If this is omitted, we obtain selective bass distortion, manifested by a peculiar hooting, rattling and tearing paper sound which is easily mistaken for a loudspeaker speech coil which has gone out of centre.

Figure 2 shows a suitable volume and frequency matching circuit, which also effects the stereo summation to derive a monaural signal for the extension speakers without introducing crosstalk in stereophonic reproduction in the living room.

## Isolating transformers

Note the audio isolating transformers in Fig. 2 and the others in Fig. 1. These are necessary to break hum loops when interconnecting the audio inputs and outputs of several pieces of equipment. Further measures necessary to avoid hum demand that the isolating transformers should not be positioned close to mains transformers from which they could otherwise pick up hum by induction, and all audio lines should be run at low impedance (normally $4 \Omega$ loudspeaker circuit impedance), so that capacitive hum pickup also remains negligible.

## Pseudo-bass

In Figure 2, C2 and C3 in conjunction with VR2 and VR3 produce the necessary bass cut to prevent overloading of the auxiliary amplifier on bass components of the full frequency range signal, whilst preserving the same subjective volume.

A peculiar property of the human ear is that it will regenerate the fundamental of a rich spectrum of harmonios, even if this fundamental is absent in the actual sound radiated by a loudspeaker. For example, the bass fundamental of a guitar is subjectively heard, even if the loudspeaker is not objectively radiating this frequency. The physiological sensation is not quite the same as when the fundamental frequency is actually present in the radiated sound, but only the musically trained ear can hear the difference in the absence of a direct comparison. Almost any normal individual can certainly hear the difference if the two
instances are played to him in quick succession, so that he will derive much more pleasure from listening to an equipment giving the full frequency range reproduction. But without musical training he tends to forget the difference when listening to other equipments with inferior frequency range, provided that harmonic distortion is low. The ear is much more sensitive to harmonic distortion, or more correctly cross-modulation distortion, in the absolute sense, although even here it is astonishingly tolerant if rhythm or speech message content of the signal are more important than harmony.

All these facts and considerations taken together mean that the reproduction from the extension speakers gives a very good impression easily mistaken for high-fidelity, after passing the full frequency range signal through the circuit of Fig. 2 and then giving it amplification with level frequency response. To the untrained ear, the difference becomes immediately apparent only when moving quickly from the extension speaker to the living room where the same programme is being reproduced with the full frequency range. Modern portable radios operate on the same physiological principle, coupled with good exploitation of cabinet and speaker resonances, to give a powerful impression of pseudo-bass from quite small units.


Close-up view of some of the amplifying equipment mounted at the bottom of the loudspeaker cabinet.

## Choice of relays

The relays in Fig. 1 should be substantial devices with the greatest possible contact pressure, especially when incorporated in one of the loudspeaker boxes. Otherwise the contact resistance can become poor and modulated when the loudspeaker is running very loud. Relay contacts in loudspeaker circuits are called upon to carry high currents at low voltages. This calls for particularly heavy contacts. Wherever available, two or more identical contact sets should be connected in parallel. If relays with mains operated solenoids are unavailable, use low-voltage a.c. types with a suitable transformer. If any type of available a.c. relay produces an audible buzz disturbing quiet listening, check whether this is merely due to a loose coil bobbin on the core assembly. If so, a few spots of Durofix or a wedge of paxolin will generally cure the trouble. If not, d.c. relays must be substituted and fed via a transformer, rectifier and reservoir capacitor. Make quite sure that all relays are rated for continuous duty, i.e. that they may be
energised for any length of time without overheating. The author has sometimes found that optimum reliability in the face of high acoustic levels inside the cabinet is obtained by using an energising current slightly below the nominal value, i.e. by insertion of a series resistor determined by trial and error. This strikes a compromise between contact pressure and elastic stability by virtue of cooler running.

## Stereo phasing

Correct phasing is important when using two loudspeaker boxes for stereophonic reproduction. The speech cones must move inwards and outwards in step when the speakers are physically and electrically identical, which is the ideal case. Identical electrical connections then correspond to the correct phasing. If physically or electrically different speaker units are used on grounds of styling or availability, very good stereophonic reproduction is still possible in most cases. Any difference in sensitivity is easily corrected with the balance control on the radio tuner. With non-identical speakers, equal electrical phasing and cone movement does not necessarily produce equal sound pressure phasing in the room. In many such cases, either electrical phasing will produce a correct stereophonic effect over certain listening areas, whereby one phasing gives only a small, usually very asymmetrically located listening area, whilst the other one gives a much larger area roughly corresponding to the normal central zone obtained with two identical speakers. There is no simple way of determining the correct case by inspection. The decision must be made in the following manner by listening.

Tune the radio to the strong hiss between two stations on the v.h.f.-f.m. band, and adjust the

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volume of this hiss to be loud and crisp. Expiore the room to determine the region in which the hiss appears to be coming from a point source on the line between the speakers. At all other points in the room, the hiss will appear to be coming from nowhere in particular, often exerting pressure on the head from above or the rear, or of course it will appear to come from one speaker if approached too closely. Now reverse the connections at one loudspeaker and repeat the exploration. The correct polarity is the one which produced the larger and more central region of sound source impression on the line between the speakers. This test must be made in the monaural setting of the radio tuner, because the decoder phasing is unreliable in the absence of a signal.

If the test does not prove decisive for either electrical phasing, try the effect of slight movement of furniture or a slight change of the mounting height of one loudspeaker.

## Combination phasing

The signals from the stereo channels fed via the circuit of Fig. 2 to the combination stage must be electrically in phase. This is easily checked by running a monaural signal from the radio tuner and noting the effect of the balance control on the volume at the extension speakers. When the phasing is correct, the balance control should have very little if any effect on the extension speaker volume. If the phasing is incorrect, the extension speaker volume will drop considerably in the central setting of the balance control. In such a case, reverse the connections to any one of the four windings of T2 and T3 in Fig. 2
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important are complete freedom from thermal runaway due to the close thermal coupling between the output transistors and the bias diodes and very low level of distortion.

The IC-10 is primarily intended as a full performance high fidelity power and pre-amplifier, for which application it only requires the addition of the usual tone and volume controls and a battery or mains power supply. However, the IC-10 is so designed that it may be used simply in many other applications including car radios, electronic organs, servo amplifiers (it is d.c. coupled throughout) etc.

The photographic masks required for producing monolithic I.C's. are expensive but once made, the circuits can be produced with complete uniformity and at very low cost. So we are able to sell the IC-10 at a price far below that of the components for a conventional amplifier of comparable power. At the same time, we give a 5 year unconditional guarantee on each IC-10 knowing that every unit will work as perfectly as the original and do so for a lifetime.

SINCLAIR RADIONICS LTD, 22 Newmarket Rd. Cambridge.Tel: 0CA3-52996

## 10 Wart Mownunul miricinili⿷匚MRUTIT AMPIIFIR

## Specifications

Power Output
10 Watts peak, 5 Watts R.M.S. continuous.
Frequency response $\quad 5 \mathrm{~Hz}$ to $100 \mathrm{KHz}=1 \mathrm{~dB}$. Total harmonic distortion Less than $1 \%$ at full output. Load impedance 3 to 15 ohms. Power gain $110 \mathrm{~dB}(100,000,000,000$ times) total. Supply voltage 8 to 18 volts.
Size
Sensitivity
Input impedance
$1 \times 0.4 \times 0.2$ inches. 5 mV .
Adjustable externally up to
2.5 M ohms for above sensitivity.

## Circuit Description

The circuit diagram of the IC-10 is shown on the right. The first three transistors are used in the pre-amp and the remaining 10 in the power amplifier. The output stage operates in class $A B$ with closely controlled quiescent current which is independent of temperature. A high level of overall negative feedback is used round both sections and the amplifier is completely free from crossover distortion at all supply voltages. Thus battery operation is eminently satisfactory.

## Construction

The monolithic I.C. chip is bonded onto a gold plated area on the heat sink bar which runs through the package. Wires are then welded between the I.C. and the tops of the pins which are also gold plated in this region. Finally the complete assembly is encapsulated in solid plastic which completely protects the circuit. The final device is so rugged that it can be dropped thirty feet on to concrete without any effect on performance. The circuit will also work perfectly at all temperatures from well below zero to above the boiling point of water.



## Applications

Each IC-10 is sold with a very comprehensive manual giving circuit and wiring diagrams for a large number of applications in addition to high fidelity uses. These include public address, loud-hailers, use in cars, inter-com., stabilised power supplies, electronic organs. oscillators, volt meters, tape recorders, solar cell amplifier, radio receivers.
The transistors in the $1 \mathrm{C}-10$ have cut off frequencies greater than 500 MHz so the preamp section can be used as an R.F. or I.F. amplifier making it possible to build complete radio receivers without any additional transistors.

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－ 28 transistor， 7 diode circuit for natural transparent sound， instant operation，long trouble－free life． 14 watts music power， 10 watts RMS from \(25-35,000 \mathrm{~Hz}\) at \(\pm 1 \mathrm{~dB}\) ．Automatic stereo indicator licht．Adjustable phase control for maximum separation． －Complete front panel controls．Flywheel tuning．All critical circuits incuding FM＂front－end＂factory assembled and aligned． －Circuit board assembly．Compact \(10 \frac{2}{⿱ ㇒ 日 勺} \mathrm{D} \times 3^{n n} \mathrm{H} \times 12^{\prime \prime} \mathrm{W}\) ． －Front panel stereo headphonə jack．
Kit K／AR－14（less cab．）．．kit £39．0．0．P．P．13／6
A／AR－14（less cab．）．．．．E59．0．0．P．P．13／6
Cabinet extra：Teak or Walnut finish £3．10．0

\section*{A Quality Table Radio FM Mono Receiver，AR－27}
－ 13 transistor， 6 diode circuit for high－fidelity sound reproduction， long life，low heat，freedom from hum，and service－free operation． 7 watts music power． \(\pm 1 \mathrm{~dB}, 25\) to \(60,000 \mathrm{~Hz}\) at 5 watts． －Input connectors for phonograph and auxiliary signals．Com－ plete front panel controls．Fiywheel tuning．Preassembled and prealigned FM tuner，all other critical parts factory aligned．Easy， circuit board assembly．Compact bookshelf size．3－way installation ．．．wall，free standing or in a suitable cabinet． 117 v ． A．C．or \(210 / 240\) v．A．C．， 50 Hz operation．
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Kit K/AR-27 (less cab.) . . kit £22,10.0. P.P. 10/6
A/AR-27 (less cab.) . . .. E32.0.0. P.P.10/6

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        Cabinet extra: Teak or Walnut finish f3.10.0


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R．S．T．VALVE MAIL ORDER CO． BLACKWOOD HALL，16a WELLFIELD ROAD，STREATHAM S．W． 16 \\ All valves \\ brand new
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\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|c|c|c|}
\hline 1 A & 7／8 & 6BH & & 6 K 6 & 787201 & 20 & & ， & & EM80 7／6 & C97 8／8 & 0 & UF89 7／8 \\
\hline 1 D 5 & 71 － & 6BJ6 & 9／－ & 6K；M 5／8 & \(7 \mathrm{Y} 4 \quad 8 / 8\) & 20P4 19／－ & 78 & 4／8 & ECH21 12／6 & EM81 7／9 & PCC84 6／3 & 801－ & L41 9／6 \\
\hline 1H5 & 71 & 6BQ7A & 71－ & \(6 \mathrm{~K} 7 \mathrm{Ca} 21-\) & \(9 \mathrm{BW6}\) 7／－ & \(20 \mathrm{P} 518 /-\) & 80 5／ & DK32 7／9 & ECH35 11－ & EM84 7／6 & PCC89 11／－ & 8130 35／－ & UL84 71－ \\
\hline \(1 \mathrm{LD5}\) & \(51-\) & 6 BR 7 & 8／6 & 6K7GT 4／6 & \(10 \mathrm{Cl} 112 / 6\) & 2546 5／9 & 85A2 713 & DK91 5／8 & ECH42 11／－ & ESU15020／－ & PCC189 11／6 & 8P4 8／ & UM80 6／－ \\
\hline 1N5GT & 81－ & 6BR8 & 5／8 & 6K8M 8／6 & 102 12／6 & 25L6GT 5／6 & 150B2 9／6 & DK92 8／－ & ECH81 5／9 & EY51 7／6 & PCF80 7／－ & 8P41 3／6 & UU6 18／6 \\
\hline 1 R 5 & 5／6 & 6Bs7 & 16／9 & 6 K 8 G 3／－ & 10Fl 9／－ & 25Y5 8／－ & \(7 / 8\) & DK96 \(7 / 8\) & ECH83 8／6 & EY86 \(7 /-\) & PCF82 6／－ & SP61 \(3 / 6\) & UU7 13／6 \\
\hline 1 S 4 & \(51-\) & 6BW & 14／－ & 6K8GT 7\％－ & \(10 \mathrm{~F} 3881-\) & 25Z4 6／3 & 801 6／－ & 151－ & ECL80 7／－ & EZ35 4／6 & PCF84 8／－ & 8TV280／80 & Uu9 8／－ \\
\hline 185 & ， & 6BW & 14／－ & 6 K 25 20／－ & 10F9 9／9 & \(25 \mathrm{Z5}\) 7／－ & 807 7／－ & DL92 \(4 / 9\) & ECL82 7\％－ & ER40 8／－ & PCF86 9／－ & 85／－ & UY21 9／8 \\
\hline 1 T 4 & & 6 C & 2／8 & 6 L 1 9／8 & 10F18 9／－ & \(25 \mathrm{Z6}\)－8／6 & 811 & DL93 \(3 / 6\) & ECL83 10／3 & EZ41 10／－ & PCF801 10／－ &  & UY严 7／－ \\
\hline 3 A 4 & 3／6 & 6C & 4／－ & 6L6G \(\quad 7 / 8\) & 10L1 8／ & 8D7 5／－ & 866A 13／6 & DL95 6／6 & ECL86 9／－ & EZ80 5／6 & PCF802 \(10 /\) &  & UY85 6／6 \\
\hline 3Q4 & B／6 & 6．66 & 3／9 & bj－ & \(10 \mathrm{Pl3} \quad 15 / 6\) & \(30 \mathrm{Cl} 513 / 6\) & \(\begin{array}{lr}8864 & 4 / 6 \\ 954 & 4 / 6\end{array}\) & DL96 \(7 / 8\) & 800 & GZ30 10／－ & PCF806 13／6 & TDD4 10／－ & MP4G17/- \\
\hline 3Q5 & 6／8 & 6CD6 & 22－ & 6Q7aT 8／6 & \(11 \mathrm{E} 342 /-\) & 30C17 \(14 /-\) &  & \(\begin{array}{ll}\text { DM70 } & 7 /- \\ \text { DY86 } & 8 /-\end{array}\) & 201 & GZ32 \(\quad 0 / 6\) & PCF80811／6 & U10 716 & \[
\mathrm{R105/30}
\] \\
\hline 384 & \(4 / 8\) & 6CH6 & 5／9 & 68A7M 7－ & 12AT6 4／6 & 30 C 18 13／6 & \(\begin{array}{ll}4022 A & 50 /- \\ 5763 & 10 \%\end{array}\) & DY86 6／－ & EF3TA 7 7－ & GZ34 11／－ & CLaz 7／9 & \(\begin{array}{lr}\text { U14 } & 7 / 6 \\ \text { U19 } & 35 /-\end{array}\) & 5／－ \\
\hline 3 V 4 & 5／9 & W & 12／－ & \({ }^{68 \mathrm{SC7}}\) 7\％－ & 12AT7 \({ }^{12 / 8}\) & 30 Fs 14／－ & \(\begin{array}{cc}5763 & 10 /- \\ 7193 & 2 /-\end{array}\) & E88CC 12／－ & F39 6／ & KT36 17／6 & PCL83 \(9 / 8\) & \(\begin{array}{ll}\text { U19 } & 35 /- \\ \text { U25 } & 13 / 8\end{array}\) & 130 \\
\hline 5 R 4 O & 8／8 & 6D6 & \(2 / 8\) & 6SG7 5／－ & 12AU6 5／9 & \(30 \mathrm{FLI} 16 /\)－ & 7475 4／－ & EA50 2／－ & 101－ & \(\begin{array}{ll}\text { KT61 } & 12 / 6 \\ \text { KT68 } & 17 / 8\end{array}\) & PCL84 7／9 & \begin{tabular}{ll} 
U26 & \(13 / 8\) \\
\hline 13
\end{tabular} & \(5 /-\)
\(15 /-\) \\
\hline & & 6 F & \(7 / 6\) & & & \(30 \mathrm{FL14}\) & A61 7／9 & EABC8071－ & 0 2／6 & T861 \(17 / 8\) & PCL86 9／－ & －゙78 3／6 & \\
\hline 3 G & & 6 F 5 C & & 6SK7GT 4／9 & 12BA6 \(6 /\) & \(30 \mathrm{LI5} 15 / 3\) & \(\begin{array}{ll}\text { ATP4 } & \text { A／3 } \\ \text { ATP5 } & 7 /-\end{array}\) & EAF42 10／－ & EF & KT81（7C5） & PENA4 201 & C191 13／－ & U111 7／6 \\
\hline \(5 \mathrm{Z4G}\) & 618 & 6 F 6 & 41 & 6SL7GT 4／9 & 12BE6 5／9 & 30 L 17 14／－ & 3／6 & EB91 3／－ & \(8 / 8\) & 151－ & PENB4 \(20 /\) & \(\begin{array}{lll}\text { U2501 } & 16 / 3\end{array}\) & U120 12／6 \\
\hline 6／30L & 131－ & 6F8G & 4／6 & 6SN7GT 4／6 & 1208GT 4／6 & 30 P 12 12／－ & \(801-\) & EBC33 \％／－ & EF89 5／－ & KT88 27／6 & PEN45 7／－ & U403 6／6 & VU508 35／－ \\
\hline \(6 \mathrm{~A}^{7}\) & 15／－ & 6 6F11 & & 6897 6／－ & 12E1 \(17 / 6\) & \(30 \mathrm{P19}\) 13／－ & AUS 718 & EBCA1 9／9 & 16 & KTW61 10／－ & PEN46 \(2 / 8\) & \(40411 / 9\) & W81M 6／－ \\
\hline 6A8G & 12／6 & 6 F 13 & 18 & GT 12／－ & 12J5GT 2／6 & \(30 \mathrm{PLJ} 15 /\) & AZ1 8／－ & EBC90 4／6 & EF92 \(2 / 6\) & KTZ41 \({ }_{\text {ML }}{ }^{6 /-}\) & PL36 10／－ & U801 23／6 & XH1．5 5／－ \\
\hline 6 AC ： & 31－ & 14 & 12／6 & \(6 \mathrm{U5G} \quad 7 / 6\) & 1257GT 7／－ & \(30 \mathrm{PL} 1315 /-\) & AZ31 9／6 & EBF80 \％ 7 & 101－ & \(\begin{array}{ll}\text { ML4 } & 17 / 6\end{array}\) & PL81 8／－ & UABC80 6 － & XPI－5 5／－ \\
\hline 6 A & 4／6 & \(6 \mathrm{~F}^{\text {23 }}\) & 13／8 & 6V6m 8／－ & 12K7GT 8 － & \(30 \mathrm{PL} 1415 /-\) & CBL31 15／－ & EBF83 8／3 & 8／6 & \(\begin{array}{ll}\text { ML6 } & 8 /- \\ \text { MSP4 } & 10 /-\end{array}\) & \(\begin{array}{ll}\text { PL82 } \\ \text { PL83 } & 7 / 3\end{array}\) & UAF42 10／3 & XPG1－510／－ \\
\hline 6 AL & \(3 /-\) & 6F＇24 & 12／－ & 6V6G \(4 / 8\) & 12K8GT \({ }^{\text {d／}}\) & 35A5 \({ }^{\text {35 }}\) & CCH35 21／－ & EBF89 816 & EF184 616 & \(\begin{array}{ll}14 & 10 / 6 \\ 7 / 8\end{array}\) & PL83 6 6－ & UBC41 \(8 / 8\) & Y63 7／6 \\
\hline & 2／6 & \(6 \mathrm{FF}^{25}\) & 12／－ & 6V6GT \(6 / 8\) & 1297GT 4／6 & \(\begin{array}{ll}35 \mathrm{Las} & 5 / 8 \\ 35 \mathrm{~W} 4 & 4 / 6\end{array}\) & C133 201－ & EBLl 14／－ & EL32 \(3 / 6\) & Y \(14{ }^{12 / 6}\) & PL84 6／9 & UBC81 8／8 & Tubes \\
\hline 6AM & 3／6 & 6F28 & 11／6 & 6X4 3／6 & 128A7 8／6 & \(35 \times 4\) 4／0 & CY30 18／3 & EBL21 11／－ & EL．33 12／6 & MX40 12／6 & PL500 15／ & UBF80 \(6 / 8\) & 3EG1 50／－ \\
\hline 6495 & 61 & & \(2 / 6\) & \(\begin{array}{ll}\text { 6X5G } & 4 / 6 \\ 6 \times 5 G T & 6 /-\end{array}\) & \(\begin{array}{ll}128 G 7 & 4 / 3 \\ 12847 & 8 /-\end{array}\) & \(35 \% 3{ }^{35 \% 4} 10 / 6\) & \(\begin{array}{ll}\text { CY31 } \\ \text { DAC32 } & 101 \\ & 7\end{array}\) & 27 & EL34 \(10 / 6\) & 108 25／ & \begin{tabular}{cr} 
PX4 & \(14 /-\) \\
PY3 & \\
\hline 16
\end{tabular} & & \(\begin{array}{ll}\text { 3FP7 } & 19 /- \\ \text { 5CP1 } & 35 /-\end{array}\) \\
\hline \(6 \mathrm{AB70}\) & \(15 /-\) & 653 M & 8／6 & \({ }^{7 \mathrm{~B} 6} 11 / 6\) & 12857 & 3525 5／6 & DAF91 4／－ & ECC81 3／9 & EL41 101－ & NGT1 3／6 & PY81 8／6 & UCC85 7－ & CV1526 40／－ \\
\hline 6 6T6 & 4／6 & & 2／6 & B7 7／6 & \(12 \mathrm{SKT} 4 / 9\) & 37 51－ & DAF96 6／9 & ECC82 \(4 / 9\) & EL42 10／－ & GGT7 55／－ & PY82 81－ & UCF80 8／6 & ACR15100／－ \\
\hline 6AU6 & 81 & \(6 \mathrm{J5G}\) & 4／8 & 7C5 15／－ & 129R7 5／－ & \(8 /\) & DCC90 7／－ & ECC83 6／3 & EL84 \(4 / 9\) & OA2 5／9 & PY83 6／6 & I＇CH42 \(10 / 6\) & VCR97 36／ \\
\hline 4 G & 151 & 6J6 & \(3 /-\) & 7 C 8 15／－ & \(14 \mathrm{H} \overline{7}\)－ \(9 /\) & 50B5 6／6 & DF333 8／－ & ECC84 61 & \(81-\) & OC3 5／－ & PY800 101－ & UCH81 8／9 & VCR517B \\
\hline 6B89 & 2／－ & 6.57 M & \(7 / 6\) & 7D5 81－ & 19AQ5 5f－ & \(50 \mathrm{C5}\) 8／3 & DF70 7／－ & ECCS5 5／ & 5／6 & OZ4 4／6 & PY801 101－ & UCL82 8／－ & 48）－ \\
\hline 6BA6 & 51－ & 6． 57 G & 4／8 & \(7 \mathrm{H7} \quad 8 / 6\) & \(20 \mathrm{D1}\) 10／ & 50CD6G81／－ & DF91 3／－ & ECC88 7／－ & ELL80 20／－ & PC86 11／6 & R2 \(2 / 6\) & UCL83 10\％－ & 17 C \\
\hline 6BE6 & 5／－ & 6.7 GT & 8／6 & \(\begin{array}{ll}7 \mathrm{R7} & 17 / 6\end{array}\) & 20 F 2 14／－ & 50L69T 6／－ & 2／6 & 6／6 & 251 & PC88 11／6 & 7／8 & UF41 101－ & \(46 \%\) \\
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\end{tabular}

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\hline AC127 & \(7 / 6\) & OC25 & 11／ & 0 C 71 & 4／6 & OC81 & & OC82D & 日） \\
\hline AFll 4 & \(7 /-\) & OC28 & 16／－ & 0 C 72 & \(81-\) & OC815） & 4／－ & \(0 \mathrm{C83}\) & 6 \\
\hline AFll5 & 71. & 0 C 35 & 11／6 & 0 C 75 & 61－ & OC8 1 DM & & 0 C 170 & 7 \\
\hline AF116 & 71 & 0 O 44 & 4／6 & OC76 & \(81-\) & & \(61-\) & OC171 & \(8 /\) \\
\hline AFl17 & 7－ & \(0 \mathrm{C45}\) & 4／－ & 0 C 77 & 8／－ & \(0 \mathrm{C8} 2\) & 81－ & OC200 & 7／6 \\
\hline
\end{tabular}

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\hline BC 108 & 3／－ & BC 113 & 3／9 & 2N 706 & 2／6 & 2N 3705 & 3／2 \\
\hline BC 109 & 3／6 & BC 182L & 4／－ & 2N 706A & 3／2 & 2N 3706 & 3／7 \\
\hline BC 115 & 10／－ & BC 183L & 3／3 & 2N 2926 & 2／6 & 2N 3707 & 3／10 \\
\hline BC 116 & 12／－ & BC 184L & 4／9 & 2N 3053 & \(61-\) & 2N 3708 & 2／6 \\
\hline BC 167 & 3／－ & BFY 50 & 5／3 & 2N 3054 & 15／6 & 2N 3709 & \(2 / 4\) \\
\hline BC 168 & 2／6 & BFY 51 & 4／4 & 2N 3055 & 21／－ & 2N 3710 & \(2 / 7\) \\
\hline BC 169 & 2／9 & BFY 52 & 5／3 & 2N 3391A & 71－ & 2N 3711 & 3／－ \\
\hline BC 170 & 3／－ & BFY 53 & 4／4 & 2N 3702 & 3／6 & 2N 3793 & 3／3 \\
\hline BC 171 & 3／8 & BSY 95A & 3／9 & 2N 3703 & 3／2 & 2N 4292 & 3／3 \\
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\(8 / 460 \mathrm{~V}\)
\(8 / 460 \mathrm{~V}\)
\(16 / 450 \mathrm{~V}\)
\(18 / 450 \mathrm{~V}\)
\(82 / 450 \mathrm{~V}\)
\(82 / 450 \mathrm{~V}\)
\(25 / 25 \mathrm{~V}\)
\(25 / 25 \mathrm{~V}\)
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