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Model TK-500E SOLID STATE FM MULTIPLEX STEREO TUNER


Anothergreat Lanky'a apecial purchase-the TK-500E is a truly outstanding 21 transistor FM Multiplex Stereo Tuner by 'TRIO-Japan's foremost producer of transistorised Hi-Fi equipment. The extremely sophisicated circuitry of the unique Thio features including Autimatic electrical awheration noise muting circuit and mono lankes with stere indicator beacon circuit eliminates secondary signal and -beat" interference even in fringe area reception Nuvistor cascode front end and 5 IF stages assure the highest sensitivity reception Pinpoint meter tuning bow impedance steren, mono and tape outputs. BRIEF SPECIFICATION: 21 transistors, 15 Germantum and 8 silicon diodes and 1 Zener diode. 3 valves. Frequency range: 88108Mc/s. Sensitivity: $0 \cdot 8 \mu A$ ( 20 dB quieting with $72 \Omega$ antenna). Output: 2 V . Frequency response: $20-20,000 \mathrm{c} / \mathrm{s} \pm \mathrm{ldB}$. Channel separation: better than 35 dB . Distortion ess than $0-6 \%$ at $400 \mathrm{c} / \mathrm{s} 100 \%$ mod. Capture ratio: 2 clB . 8 uperbly styled and inished hammer enamel and brushed alloy cabinet, size $15 \frac{1}{4} \times 12$
operation. Today's comparative value over 865 .
Lasky's Price 38 Gns. Carriage and Packing 12/6.

## TRIO

Model TK-150T SOLID STATE STEREO AMPLIFIER


#### Abstract

Trin equipment is refor quality-now this amous eompany break the price barrier new budget priced Hi-Fi unit. The TK. Hi.Fi unit. The Th. compact 19 transistor and 8 diode stereo mplifier giving 40 watts music power, 13W HMB power per provided for Magnetic  pick-up ( $2 \cdot 1 \mathrm{mV}$ ). Tuner ( 130 mV ), and 2 Auxiliary Inputs ( 130 mV each) for use with another Tuner or Tape Recorder, sep. input for tape recorder ( 130 nV ). Built-in tape monitoring circuit. Outputs tor speakers, stereo headphones, tape play, 00 A.C. power outlet also provided. Controls include: Volume ( $L \& R$ ), bass, treble, input selector power on/off, loudness, mode (stereotmono), tape moner play the last  better than $65 d \mathrm{~B}$, Tuner/Aux. 1 and 2 -better than RTAA equalisation, Built -ith power ransistor protection circuit. Power mily $10!\times 98 \times$ 4 ia. leark matit finish control panel with silver anodised trim and black/silver controls Complete with detalled instruction manual and circuit data. Lasky's Price $£ 35$ Carriage \& Packing 8/6.


HI-FI BOOKSHELF SPEAKER BARGAIN-FOSTER FCS-166 This extremely high quality bookshelf speaker system by the worla wopfer and cone tweeter in a sealed inflite baffe enclosure with handsome oiled walnut finish. The performance of the FCS-166 is superior to many larger and far more expensive unita and at Lasky's epecisl purchase price is quite without equal! SPECIFICATION: Air suspension type $6 \frac{1}{i n}$. bass-midrange wooter with rolled cloth edge. tim. 1fF cone tio tweeter. on impedance. Cabinet construce ted from sin laminate with oiled wainut veneer Anish; size $139 \times$ ain. square. Dark green woven acoustic ganze. Phono input at rear.
Lasky's Price £9.19.6


Lasky's Price for 2-£18

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Lasky's money saving Package Deal Plan guarantees you a substantial cash saving when you buy a complete system plus the assurance that each item has been carefully chosen for comprising equipment on this page.
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TRIO TK-500E FM \&tereo Tuner . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 838.18 .0
TRIO TK-150T Stereo Amplifier. . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 885.0 .0
2 FOSTER FCS-166 Speaker systems . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . $\quad$ £18.19.0
GARRARDAT.60 Mk II 4 speed autochanger. . . . . . . . . . . . . . . . . . . . . . . . . . . . $\quad \$ 18.19 .8$
Base and cover for AT. $60 . . . . . . . . . . . . . . . . . . . . ~$
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$301-$
H.P. Terms: \&26.10.0 dep. 12 monthly payments of 87.9.1. Total H.P.P. \&15.18.8 PACKAGE DEAL B

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Lasky's Package Price $£ 72 \underset{\substack{\text { c. } \\ 30 /-}}{ }$
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switch for all ranges. Ohms zero adjustment switch for all ranges. Ohms zero adits $0-6 \cdot 30-300-1200 \mathrm{~V}$ a Range spec.
$10 \mathrm{~K} / \mathrm{ohma} / \mathrm{V}$. volts: $0-6 \cdot 30-300-1200 \mathrm{~V}$ volts: $0-3-15-150-300-1 \cdot 2 \mathrm{KV}$ at $20 \mathrm{~K} /$ ohms $/ \mathrm{V}$. Resistance: $0-60 \mathrm{~K}-6 \mathrm{megs}$. DC current: 0-60 LLA-300mA. Decibels: $-20 \mathrm{dl3}$ to +17 lB . Hand calibration given extremely high standard of accuracy on all ranges. Uses
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responpe $50-10 \mathrm{ke} / \mathrm{s}$, distortion $3 \%$ at $200 \mathrm{n} W$. responae $50-10 \mathrm{ke} / \mathrm{s}, \mathrm{distortion} 3 \%$ at 200 nuw .
$\mathrm{E}-1815$ Eleotronic Organ (tone oscillator) Module-irequency $200-1000 \mathrm{c} / \mathrm{s}$, output 80 mW For use with teyboard, variable resistors and $8 \Omega$ speaker. $\mathrm{c} / \mathrm{s}$, output 80 PRICE $25 /$ E-1316 Morse Code Practice Oscillstor Module-frequency $400 \mathrm{c} / \mathrm{s}$, output 80 mW . For use
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nel $\star$ Separate Bass and Treble con－
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 2.5/10/50/250/07.50 $1,000 \mathrm{~V}$ IB.O. $0 / 3 / 10 / 50 /$
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## PROFESSION-

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(per channel)
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Employs Mullard valves throughout. ECC83 and $2 \times$ ECL86 with a metal bridge rectification.
TECHNICAL SPECIFICATIONS Gram sensitivity 40 mV at 1 KHz . Aux sensitivity 50 mV at 1 KHz . (Sensitivities are given for rated output). 4 watts r.m.s.
( 8 watts r.m.s. In monaural position) Output matches ino standard 3 ohms speaker system. Suitable $10^{\prime \prime} \times 6^{\prime \prime}$ speakers are avail-
able at $29 / 6$ each, plus $5 /-$ p. \& p. Bass control at 100 Hz lift +9 dB 10 KHz lift +8 dreble control at Total +8 dB cut -13 dB . t 3 watts and $2 \%$ distortion $0.35 \%$ at 1 KHz . Negative rated output at 1 KHz . Negative feedback 13 dB at 1 KHz. Mains supply $220-250 \mathrm{~V}$
A.C. $50-60 \mathrm{~Hz}$.

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Inputa 10 watta RMS Music-power
Tone Controls-Treble control range $\pm 12 \mathrm{~dB}$ at 10 KH 2 Frequency Rram/radio 250 mV Bass control range $\pm 13 \mathrm{~dB}$ at $100 \mathrm{H2}$ Bignai to Noise Ratio-better than-60dB. Transistors-4 ailioon Planar type and 3 Germanium type. Mains input- $220-250 \mathrm{~V}$. A.C. Size of chassis- $10^{\circ} \times 32^{\prime \prime} \times 2^{\prime \prime}$. A.C. Mains, $200-250$ V. For use with Std or L. P. records, musical Instruments, all makes of pick-ups and mikes. Separate bass and treble lift control. Two inputs with control for gram.


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R.M.s. Power Output: 13 W (music power), 10 W (Sine Wave) sensitivity: for rated output 1 mV into 3 K ohms load.
Total Distortion: at 1 KHz for rated output 1.50 Hz and 40 KHz .
$0-35 \%$.
Outint Imperiance: $3 \mathrm{ohms}(3-15$ ohms may be used)
Supply Voltace: 24 V D.C. at 800 mA ( $6-24$ V may be used) output at Size:24* $\mathrm{K} 3^{-} \times 1^{9}$. supply with 3 ohms speaker. 7 watts
ize full ${ }^{16}$
The fully comprehensive instruction manual does not only show the practical easy-to-understand detailed information about the Xives. Standard equalisation networks are given for most types of conventional inputs. They include: Tape Head, Mag. P.U., Xtal P.U., Tuner Mic, etc. $49 / 6+2 / 6 \mathrm{p}$ \& p .

Control hssumbly: (Including resistors and capacitors) 1. Volume: PRICE 5/-. 2. Treble: PRICE 5/-

The above 3 items can be purchased for use with the X101.
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## [D] 1$] / 1$ a high quality monaubal <br> PRE-AMP \& CONTROL UNIT

Particularly suitable for use with the X101 if a ready-built. comprehensive, multi-input system is desired.

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Selector Switch. Tape Speed Equalisation Switch (37 and 7 i.p.s.), Volume. Treble, Bass, 3 position scratch fitter and position rumble filter.

## SPECIFICATION

Sensitivities for 200 mV output at 1 KHz
Tape Head: 3 mV (at 3i i.p.s.) Radio: 100 mV
Mag. Pu.: $\quad 2 \mathrm{mv}$
Tapertere Ontput: 100 mV
Equalisation for each input is correct to within $\pm 2 \mathrm{~dB}$ (R.I.A.A.) from 20 Hz to 20 KHz .
Tone Control ltangl: Bass $\pm 13 \mathrm{~dB}$ at 60 Hz
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Sifnal Noise: $>-60 \mathrm{~dB}$
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## Th ACQB HIGH qUALITY SOLID-STATE AMPLIFIER (MONO)

## SPECIFICATION

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\section*{TOPIC ロF THE MONTH}

\section*{The Sound Scene}

MANY "radio enthusiasts" are interested in subjects not strictly speaking radio at all, such as electronic devices, test equipment and the type of interest referred to collectively as "audio".

The last decade has seen a remarkable growth in audio interest. At one time, an amplifier was a relatively simple proposition involving perhaps two or three valves in one of several conventional configurations, and associated equipment was similarly fairly "standard". Since then, great strides have been made in the design of components and circuitry, helped along by several innovations in recording and playback techniques.

First came microgroove disc recordings, with attendant raising of standards in playback equipment, then came the rapid development of tape recording and now we have the introduction of stereo on disc and tape recordings and for f.m. radio broadcasting.

During the time that audio has been flourishing, a multitude of specialist shops have sprung up and even many ordinary radio and TV dealers have made an effort to establish departments capable of handling and exploiting today's sophisticated audio products.

Also, in that time, the spread of participation has grown accordingly and an audio enthusiast today may design his own equipment, build it from P.W. articles, make it up from kits, or buy build-up units and interconnect them into a system. He may, in fact, be anything from a keen technician to a complete layman.

A result of this audio explosion has been a bewildering proliferation of products-commendable in that the prospective buyer has a wide choice but an inbuilt disadvantage because of the confusing variety available.

Our new series of special supplements is aimed to help readers sort out the wheat from the chaff, to provide a concise but informative survey of what to look for in the constituent units in a modern audio complex. The supplements will be useful even to those who normally build their main equipment, for certain items-such as loudspeakers, deck mechanisms and microphones-must be purchased.

This month we cover the main units of an audio set-up, next month we deal with the end links, the transducers, and finally we will cover workshop practices. Plus, next month, a special wall chart of reference material.
W. N. STEVENS-Editor.

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\footnotetext{
"With this month's issue the price of PRACTICAL WIRELESS has been increased to \(3 /-\). This is the first increase since May 1966. Since that time all costs have increased and the position has now been reached when we must ask our readers to make some contribution if the standard and authority that has been associated with PRACTICAL WIRELESS from its foundation is to be maintained. With the next issue we change our production methods to web-offset and this will permit better reproduction of photographs which will be of great benefit in constructional articles.
We are sure our readers will appreciate this improvement in production and will understand the reasons for the increase in price."
}
nOVEMBER ISSUE WILL be published on OCTOBER 4th

\footnotetext{
All correspondence intended for the Editor should be addressed to : The Editor, "Practical Wireless'", George Newnes Ltd., Tower House, Southampton Street, London, W.C.2. Phone: TEMple Bar 4363. Telegrams: Newnes Rand London. Subscription rates, including postage: 42s. per year to any part of the world. (C) George Newnes Ltd., 1968. Copyright in all drawings, photographs and artlcles published in "Practical Wireless' is specifically reserved throughout the countries signatory to the Berne Convention and the U.S.A. Reproductions or imitations of any of these are therefore expressiy forbidden.
}

\section*{MICRO-MASSIVE}

Microelectronics, small as the resultant units are, is spreading. Marconi-Elliott have recently opened what they claim is Britain's largest and most modern micro-electronics plant, at Witham in Essex. Personnel already approaches the 450 mark.

The standard of air cleanliness in production areas is controlled to very precise limits. Dust concentrations in the air are filtered out conforming to Class 1,000 conditions-no more than 1,000 particles greater than 0.5 of a micron in diameter in every cubic foot. Typical figures for air in the average office could be 500,000 particles up to 250 microns in diameter.

After primary filtration, the air passes through a refrigerator bank circulating 270 tons of refrigerant per hour which removes the water vapour. It is then split. part being raised in temperature by a heater supplying over \(2 \frac{1}{2}\) million B.Th.U's per hour. Any area is then controlled in temperature by blending the two supplies. Some areas have even cleaner air -Class 100! (100 particles less than 0.5 microns in diameter per cubic foot).

ELECTRONIC SLIDE RULE


The HP 9100A is a programmable, electronic calculator which performs operations commonly encountered in scientific and engineering problems. The easily-readable cathode ray tube instantly displays entries, answers and intermediate results. Not much larger than an office typewriter, this midget mathematical brain will perform a wide variety of intricate mathematical functions. It will read up to 10 significant digits with automatic decimal point placement.

Operations appear almost endless and includeaddition, subtraction, multiplication, division and finding the square root. Other operations include- \(\log x ; \ln x\) and \(e^{x} ; \sin x ; \cos x ; \tan x ; \sin ^{-1} x ; \cos ^{-1} x\) and \(\tan ^{-1} x(x\) in radians or degrees). Also available are the hyperbolic functions \(-\sinh x, \cosh x, \tanh x, \sinh ^{-1} x, \cosh ^{-1} x\) and \(\tanh ^{-1} x\). In addition, the calculator will provide coordinate transformation from polar to rectangular, rectangular to polar, cumulative addition and subtraction of vectors. The unit can be programmed and this can be recorded on small magnetic cards, the capacity being 196 steps. Further details are available from the makers, Hewlett-Packard, 224 Bath Road, Slough.

AUDIO EQUALISER


Leevers-Rich have introduced a new audio equaliser. model A501. which covers the audio spectrum in seven separately adjustable, overlapping. logarithmically spaced bands. Control knobs for the seven "constant B" type equalisers are arranged on the front panel to provide a graphic display of the correction applied.

Features of the A501 include negligible distortion, low noise, zero insertion loss, switchable HP and LP filters and a wide control range. The equipment is all solid state.

Specifications are: Input-600 ohms, bridging or terminating, balanced or unbalanced: Gain-unity, adjustable \(\pm 10 \mathrm{~dB}\); Output-isolated, to feed 600 ohm load at +20 dBM max: Frequency range \(-30 \mathrm{c} / \mathrm{s}\) to \(20 \mathrm{kc} / \mathrm{s}\) \(\pm 2 \mathrm{~dB}\) : Noise level-below - -60 dBM ; Centre frequencies - 40, 100, 250, 630. 1.600, 4,000 and 10,000 cycles: Control range- +8 dB to -8 dB (each band): HP filter—Off \(/ 70 \mathrm{c} / \mathrm{s} / 100 \mathrm{c} / \mathrm{s}\); LP filter— \(7,000 \mathrm{c} / \mathrm{s} / 10,000 \mathrm{c} / \mathrm{s} /\) off; Low level switch-reduces both maximum output and noise level by 20 dB , other parameters not affected.

Further details may be obtained from Leevers-Rich Equipment Ltd., 319 Trinity Road. London. S.W. 18.

\section*{WAFER SPEAKERS}


Oakland Trading Company, announce the new PolyPlanar "Wafer-Type" wide-range electro-dynamic speakers in a 5 watt model. The unit is extremely thin, about one-fifth the depth of conventional cone speakers in the same power range. Weighing only 11 oz . the P5 shown measures only \(8 \frac{1}{2} \times 4 \frac{1}{2} \times \frac{13}{16} \mathrm{in}\). and is made of expandable polystyrene plastic.

Further details are available from the Oakland Trading Company, 68 Lupus Street, London S.W.1.

\section*{R.F. BOMB}

A safe-area power rating of 100 watts and an \(F_{T}\) of \(30 \mathrm{Mc} / \mathrm{s}\) are features of the latest discrete-emitter power transistor from SGS-Fairchild. They are designating this little r.f. bomb the BLY72.

The transistor has 262 discrete-emitter sites and a diffused channel stopper. In addition to its 100 watt safe area power rating it features high voltage ( 60 V minimum LVCEO), high current ( \(\mathrm{V}_{\text {CE }}\) sat. \(=1.7\) volts at 10 Amps. Encapsulation is in a TO-61 isolated collector stud package.

\section*{HOBBIES MANUAL}

Electroniques have just published the second edition of their Hobbies Manual. This new, enlarged and entirely revised 1968 edition contains 960 Pages and costs 16 s .6 d . although this cost is offset by vouchers valued at \(£ 25\). It comprises a vast catalogue of goods from single components to complete kits, in fact just about everything the home constructor is likely to need for almost any project. The Manual is available from Electroniques dealers or from Electroniques, Edinburgh Way, Harlow, Essex.

OWL OPTICS


Seeing in the dark without the aid of a source of infra-red radiation to illuminate the object or scene being observed is now possible using an image intensifier tube say Mullard.

The tube operates by using a wide diameter objective lens to collect up as much light as possible reflected by the object or scene being observed. This light is then focussed on to a photoemissive surface (known as a photocathode) which converts it into electrons. These electrons are then directed and greatly accelerated by means of metal plates, connected to a high positive voltage on to a phosphor screen. Because of their very high velocity they cause more photons to be emitted from this screen than were first received by the first photocathode. hence the original "image" is "intensified".

To ensure that as much of the light as possible is transferred from the phosphor screen of the first tube to the input photocathode of the second tube (and so on) it is guided by means of special transparent fibres (fibre optics). finally ending up as a visible image on a miniature "television type" screen 25 mm in diameter. The sensitivity of the tube makes it possible to see clearly and recognise individuals and objects under starlight conditions.

The present viewer uses three tubes coupled by fibre optics and requires a total e.h.t. of 45 kV at a minute current.

HI-FI EAGLE

B. Adler \& Sons (Radio) Ltd., 32a Coptic Street, London W.C.1. have sent the following specifications of the new Eagle RA. 96 tuner amplifier.

Tuner section: \(88-108 \mathrm{Mc} / \mathrm{s}\) with a sensitivity of \(2 \mu \mathrm{~V}\) for \(20 \mathrm{~dB} \mathrm{~s} / \mathrm{n}\) ratio and image rejection better than 55 dB . F.M. stereo separation better than 28 dB at \(1 \mathrm{kc} / \mathrm{s}\). The a.m. range covers \(535-1,605 \mathrm{kc} / \mathrm{s}\) with a sensitivity of \(700 \mu \mathrm{~V}\) and image rejection of 35 dB .

The amplifier section has an output of 20 watts r.m.s. (10 watts r.m.s. per channel) with less than \(1 \%\) distortion measured at 8 watts and a response of 20 to \(20,000 \mathrm{c} / \mathrm{s}\) \(\pm 2 \mathrm{~dB}\). Five inputs are available: Magnetic, Ceramic, Aux, Tape monitor and Tape recorder output. Tone controls are bass- \(\pm 12 \mathrm{~dB}\) at \(50 \mathrm{c} / \mathrm{s}\) and treble- \(\pm 12 \mathrm{~dB}\) at \(10 \mathrm{kc} / \mathrm{s}\). A low-cut filter provides -10 dB at \(50 \mathrm{c} / \mathrm{s}\), while the high-cut filter allows -13 dB at \(10 \mathrm{kc} / \mathrm{s}\). Channel separation is given as better than 45 dB .

The circuit uses 2 integrated circuits (preamp), an f.e.t. front end (f.m.), 24 transistors and 16 diodes. Size is \(14 \times 11 \frac{5}{8} \times 5 \mathrm{in}\). and the price is \(£ 86\) which includes purchase tax of \(£ 172 \mathrm{~s}\).

NEW P.S.U.


A new improved version of the Heathkit transistorised, regulated low voltage power supply, model IP-27, is offered by Heathkit at a price of \(\mathbf{£ 4 6 1 2 \mathrm { s } \text { or }}\) assembled at \(£ 55\) p \& p 9s. With an input of \(120 / 240 \mathrm{~V}\) a.c. at \(50 / 60 \mathrm{c} / \mathrm{s}\) an output of \(0.5-50\) volts d.c. is available with better than \(\pm 1.5 \mathrm{mV}\) regulation from zero to full load. There are four current ranges\(50 \mathrm{~mA}, 150 \mathrm{~mA}, 500 \mathrm{~mA}\) and 1.5 Amps. The unit is entirely solid state with an adjustable current limiter for 30 to \(100 \%\) on all ranges and is fully portable.


\section*{amplifier module}

\section*{R. LEYLAND}

THIS mains-powered a.f. amplifier is constructed on a small Veroboard panel measuring approximately \(2 \frac{1}{2} \times 3 \frac{3}{4} \mathrm{in}\). At each end, a heat sink gives protection to the components and also serves as a support during soldering. Components are mounted upright to occupy the minimum area, and hole spacing on the Veroboard accommodates capacitor lead spacing satisfactorily. Flat mounting of components would secure them more firmly, but a board of larger area would be required.

An output on speech and music approaching 3 watts is obtained in a high flux \(15 \Omega\) loudspeaker. When a continuous sine wave test signal is applied, more current is drawn from the supply and there is an appreciable drop in voltage. This restricts the output available under test conditions to about 2 watts. A loudspeaker of less than \(15 \Omega\) impedance should not be connected as the currents in the output transistors to provide the increased power output would be excessive and would cause overheating and damage to these transistors.

\section*{Circuit Description}

Included on the circuit board, in addition to the power amplifier, is a two-stage preamplifier with provision for connecting a treble control. The coupling capacitor, C6, between the pre-amplifier and power amplifier allows separation of the d.c. conditions in the two circuits. It is arranged that the collector voltage of Tr 2 is at a lower voltage than at the base of \(\operatorname{Tr} 3\), so ensuring the necessary polarizing voltage for the electrolytic capacitor, C6.

The mid-point voltage at the output is closely maintained by \(\operatorname{Tr} 3\), to a value set by the potential divider consisting of the resistors R9, R10, R11 and R4 in series across the 24 -volt supply. Carbon film resistors of \(5 \%\) tolerance are used throughout the amplifier, making pre-set adjustments unnecessary, except for the choosing of a suitable value for R17.

The d.c. feedback loop via R14 also stabilises the currents in \(\operatorname{Tr} 3\) and \(\operatorname{Tr} 4\). The current in \(\operatorname{Tr} 3\) is determined by the base-emitter potential of Tr 4 across


R12 and will be under 1 mA . This current also flows in Tr5 and forms part of the quiescent current of this transistor. The current in the driver transistor, Tr 4 , is the amount required to drop half the supply voltage across the total resistance in the collector circuit. Because the driver transistor is a Class A stage, its quiescent current is much larger than the quiescent current in the Class B output stage.
Even with negative feedback, it is only possible to eradicate every trace of crossover distortion in the output stage by providing a small quiescent current of about 3 mA . This is increased in Tr5 by an additional 1mA flowing from Tr 3 to a total of about 4 mA . The amount of this quiescent current is adjusted by changing the value of R17 as already described. Variation of the quiescent current with temperature is compensated by a Varite thermistor, type VA1077. This thermistor is rectangular in shape and measures \(7.5 \times 18 \times 1.5\) millimetres. It is colour-coded black, red, orange, and is the standard component for the circuit, which is on conventional lines in the driver and output stages.
The bootstrap capacitor, CII, applies the output voltage to the upper end of the driver load resistance R16, and providing the value is restricted to \(25 \mu \mathrm{~F}\), permits a large drive voltage to be obtained with less inherent distortion.

Anything short of severe mismatching in the output stage will tend to be disguised by the action of the negative feedback. Nevertheless a matched complementary pair of transistors is necessary for optimum performance. These transistors, p-n-p type AC128 and n-p-n type AC176 must be put in their correct positions, otherwise the transistors will be damaged. Special care should be taken as the transistors look alike, and the markings can no longer be seen when the transistors are in the cooling clips.

Emitter resistors of \(2 \cdot 2 \Omega\) are required with the output transistors to improve thermal stability. Resistors of this value can be obtained, but the majority of suppliers only have values down to \(4 \cdot 7 \Omega\). Two \(4 \cdot 7 \Omega\)
resistors are therefore used in parallel at each emitter, giving a slightly higher effective value of \(2 \cdot 35 \Omega\).

The driver transistor, Tr4, dissipates 240 milliwatts, and is mounted in a cooling clip on the same heat sink as Tr6. Preceding the driver transistor is an n-p-n transistor type AC127, and this should be enclosed in an insulating sleeve, since in this transistor, the collector is joined to the metal case.

The preamplifier, consisting of the transistors Tr 1 and \(\operatorname{Tr} 2\) increases sensitivity to about 8 millivolts r.m.s. for maximum output on a sine wave input signal. Decoupling of the two stages by R5 and C3 prevents undesirable feedback via the supply. Direct coupling between the preamplifier stages enables d.c. stabilising feedback to be applied over the two stages via R1. Negative signal feedback is applied through R11 to R4 in the emitter circuit of Tr1. The capacitor C3 reduces high-frequency distortion and noise. Additional frequency-dependent feedback can be applied by means of an external \(100 \mathrm{k} \Omega\) treble control. At maximum treble, the h.f. response extends to -3 db at \(15 \mathrm{kc} / \mathrm{s}\), and at minimum, it is about 14 db down at \(10 \mathrm{kc} / \mathrm{s}\).

The bass response extends to -30 dB at \(60 \mathrm{c} / \mathrm{s}\), some attention being provided by the low value of C12. If loudspeaker resonance occurs at about \(60 \mathrm{c} / \mathrm{s}\), the bass response may be increased, but only slightly, because of the low output impedance. Capacitor C12 is kept outside the main feedback loop via R14, since low capacitance values in the feedback loop would tend to produce a peak in the low-frequency response. One advantage of the \(200 \mu \mathrm{~F}\) capacitor is that it takes up a minimum of space, and its working voltage need only be greater than half the supply voltage.

\section*{Construction}

To improve heat transfer between the transistors and the cooling clips, silicone grease is applied before inserting the transistors. Burrs should be removed at holes when the cooling clips are to be fastened. As neither the heat sink nor the cooling clip are perfectly flat, it is best to smear some silicone grease on the cooling clips to fill any minute air gaps existing between them, air being a poor conductor of heat. The heat sinks are painted matt black except for the area occupied by the cooling clips, and a margin of \(1 / 16 \mathrm{in}\). around them. Electrical connection is made to the heat sinks on the flanges, and a small area around the securing bolts can be left unpainted for the purpose.

The leads of the output and driver transistors should be straightened to avoid contact with the heat sink or with each other. The mode in which
the transistors are inserted into the cooling clips is shown in the diagrams of Fig. 3.
At some positions on the Veroboard, two jumper links have to be inserted into the same hole. This is possible by using 24 s.w.g. tinned copper wire for the jumper links. When the hole has been located on the Veroboard, the wire is passed through from the copper side and back through the other hole where it is bent and held while the wire is drawn tight and pressed flat. The bent portions on the copper side are clipped with side cutters, leaving enough for soldering. If the bends are in opposite directions, the jumper link will stay in place until it is soldered. The jumper link adjacent to R11 should be insulated; similarly the lead of R12 adjacent to C10. All electrolytic capacitors should be fully insulated, either by a sleeve or if necessary by wrapping them with plastic tape before putting them into the circuit.


Fig. 3: Details of the heat sinks and transistor connections.
Efficient soldering depends upon the quick transference of heat, so that the temperature rise is localized. The tip of an instrument-type soldering iron is filed flat and tinned, and requires to be at the right temperature. With sufficient downward pressure on a flat joint, the 18 s.w.g. printed circuit solder melts immediately and can be fed in to produce a good soldered joint in less than a second. This can be recognised by the absence of gaps or irregularities. The solder should have flowed evenly all over the component lead (which must not be moved when the solder is setting) and spread on the metal to produce a bright smooth dome of solder. This method is ideal for transistor leads which, in common with the jumper links, are bent over and clipped to leave enough for soldering. The transistors are inserted after the soldering of the other components is completed, and heat sink tweezers are applied on the particular transistors lead being soldered. The length of the leads of the output transistors is determined by the position of the cooling clips on the heat sinks. The height of the other transistors can be made just enough to allow a cooling clip, held vertically, to slide underneath the transistor. In making connections to the amplifier, the soldering iron should not be allowed to come against any of the transistors.
The possibility of inserting components into the wrong positions, and later needing to remove them,


Fig. 4: Circuit of a suitable mains power supply.
encourages the use of another method for the resistors and capacitors, with their thicker leads. This is to keep the component leads straight, clip them to length, and hold the component in position until one lead is temporarily secured with solder. Solder is then run in around the leads. A clearance of about \(\frac{1}{8}\) in. between components and the circuit board is usually advisable and allows the use of heat sink tweezers. It is not so quick and easy a method as the other one, and the soldering iron is only applied for a moment at a time, allowing the joint to cool before any further attempt is made.

Resistors are prepared for upright mounting by bending one lead over some form of \(1 / 16 \mathrm{in}\). diameter mandrel. Capacitors similarly have the negative lead brought down their sides, preferably inside the insulating sleeves. Components should be inserted as shown in Fig. 2.

\section*{Test Procedure}

By following a cautious procedure in trying out the amplifier for the first time, it is possible to discover any mistake before damage is caused. There is only a narrow margin of safety with germanium output transistors, and small types are particularly susceptible to failure from thermal runaway. It is inadvisable to switch on without making some preliminary adjustments and measurements. A shortcircuit or open-circuit if present in the amplifier is liable to cause immediate damage.

Apart from errors such as the omission of a jumper link, or wrong placing of components. the closely-spaced copper strips could become bridged with solder at some point. For example, solder could fall from the tip of the soldering iron without being. noticed. The careless use of test prods or similar methods of making temporary connection is not recommended and steps should be taken to see that the amplifier cannot come into accidentay contact with metal objects.

A check is first made on the voltage of the power supply unit before connecting the amplifier: It should not exceed 26.5 volts and will be less when the amplifier is connected and drawing current. It is necessary to measure the current taken by the output transistors on no signal, while gradually increasing the voltage from zero up to its full value. This can be done by connecting the amplifier to the power unit via a suitable wire-wound potentiometer of \(1,000 \Omega\). A value has to be selected for resistor R17 to give a current of between 4 and 5 mA in Tr5. This current is measured by omitting the link wire between the terminal pin at the collector of Tr 5 and the tag
—continued on page 407

\title{
transistorised CALIBRATION ostillation
}

AN invaluable aid in any amateur or professional electronics workshop is an accurate frequency calibration standard, this device being essential for the accurate calibration of oscilloscopes, audio generators, radio receivers, etc. Such a device should have a frequency accuracy at least ten times greater than that of the instrument to be calibrated, and an absolute accuracy of better than \(0.1 \%\) will generally prove to be more than adequate in most applications. Ideally, the unit should be portable, reasonably inexpensive, and easy to use.

The unit described meets all of these requirements, consisting of a crystal controlled oscillator, followed by two decade frequency dividers, the device giving switch-selected output frequencies of \(100 \mathrm{kc} / \mathrm{s}, 10 \mathrm{kc} / \mathrm{s}\), and \(1 \mathrm{kc} / \mathrm{s}\). A seven transistor circuit is used. The amplitude of the output signal can be varied by means of a built-in attenuator.

\section*{How it Works}

The full circuit diagram of the unit is shown in Fig. 1. The crystal oscillator section is made up by Tr 1 and Tr 2 . Basically, Tr 1 is wired as a grounded base amplifier, with its base-bias fixed by R1 and R2 and decoupled by Cl ; the input signal is applied to the emitter of such an amplifier, and the output taken from the collector. If a positive-going signal is applied to the emitter, the emitter-base potential

\section*{F.L.THURSTON}
of the circuit will be increased, and the collector current will rise, so increasing the potential drop across the collector load, R3, and causing the collector to move in a positive direction. Thus, the input and output signals can be seen to be in phase.

Transistor Tr2, on the other hand, is wired as an emitter follower, with its base wired to Tr 1 collector. This transistor also gives an output, from its emitter, which is in phase with the input signal to its base, but this circuit gives an impedance transformation, with unity voltage gain. Hence, it can be seen that Tr 1 gives voltage gain, and the output signal at Tr2 emitter is in phase with the input at Tr emitter; these are the essential requirements for oscillation, and the circuit can be made to oscillate by simply coupling the two emitters together via a blocking capacitor.

In the circuit diagram, this coupling is achieved via the crystal, and, since the crystal represents a very low impedance at its designed operating frequency, the circuit oscillates at the frequency of the crystal with which it is used. The output signal can be taken, at low impedance, from Tr2 emitter.

Since exceptionally high orders of accuracy are not


Fig. 1: Complete circuit of the calibration oscillator.
required from this circuit, there is no need to take measures to counteract the slight shift in phase characteristics that occurs in the transistors at the \(100 \mathrm{kc} / \mathrm{s}\) operating frequency, and, as it stands, the operating frequency will be accurate to considerably better than \(0 \cdot 1 \%\).

The \(10 \mathrm{kc} / \mathrm{s}\) frequency standard is obtained from the \(\operatorname{Tr} 3-\operatorname{Tr} 4\) astable multivibrator, which is synchronised to the crystal oscillator by means of sync pulses fed to \(\operatorname{Tr} 3\) base via C3. This synchronising action is best understood with reference to Fig. 2a, which shows the waveforms involved in the circuit. This diagram shows that, in the conventional free-running version of the astable multivibrator, the base of Tr 3 is biased in a positive direction, thus reverse biasing the emitter-base junction, when Tr 3 is off. This positive bias decays exponentially, and as soon as the


Fig. 2a: The effect of the synchronising pulses and "locking action" on the main waveform.
emitter-base junction again becomes forward biased the transistor begins to conduct and the circuit "flips" or changes state. In the triggered version of the circuit, on the other hand, brief negative-going synchronisation pulses are imposed on the base waveform, and, as indicated in the diagram, these bring Tr 3 into conduction prematurely and trigger the circuit. Thus, the operating frequency of the multivibrator is "locked" to the frequency of the trigger signal, and accurate frequency division is achieved.

In this particular circuit, the synchronising signal is applied to one transistor ( Tr 3 ) only, and the output


General view of the completed instrument.
waveform has an uneven mark/space ratio: this is of little importance in most applications, however, and has the advantage of giving very stable operation.

The second frequency divider stage, Tr5-Tr6, on the other hand, has the synchronisation signal applied to both transistors. Here, the two transistors share a common emitter resistor, R14, across which the sync signal is applied. This sync signal is derived from the output of the \(10 \mathrm{kc} / \mathrm{s}\) multivibrator, and is differentiated by C6-R14 and discriminated by D1, to give a final series of unidirectional pulses of short duration, which synchronise the multivibrator in a manner similar to that already outlined.

The main disadvantage of this method of operation, which is essential if a \(1: 1\) mark/space output waveform is required, is that the two time constants of the astable circuit must be fairly closely matched if stable operation is to be achieved. Satisfactory operation can be achieved, however, by using fairly close tolerance values of timing components and taking reasonable care in initial setting up of the division frequency. The prototype unit gives stable operation even when the supply potential is reduced from 9 to 3 volts.

The final stage of the circuit is the emitter follower output transistor, \(\operatorname{Tr} 7\), which uses variable resistor VR3 as its emitter load. The input to Tr 7 is selected by Sla, and may be either \(100 \mathrm{kc} / \mathrm{s}\) from Tr2 emitter, \(10 \mathrm{kc} / \mathrm{s}\) from Tr 4 collector, or \(1 \mathrm{kc} / \mathrm{s}\) from Tr 6 col-


Fig. 2b: Veroboard connections and layout of components.
lector, and the final output signal is taken from VR3 slider and fed, via C9, to the output socket, SK1. The built-in 9 volt battery supply is connected to the circuit via Slb.

\section*{Construction}

For ease of construction, the major part of the electronic circuitry is wired up on a small piece of Veroboard panel, thus retaining all of the advantages of printed circuit construction, while involving none of the complications of marking out, etching, etc., which are normally involved in printed circuit practice.


Start construction by cutting the Veroboard panel to size, as shown in Fig. 2b, and then break the copper strips, with the aid of a small drill or the special cutting tool that is available, as indicated. Now drill the two small mounting holes, to clear 6BA screws, where shown, and cut back the copper around them to eliminate any danger of shortcircuits when the panel is finally secured in place on the main chassis.

Now assemble the components and leads on the plain side of the panel, as shown in Fig 2b, and solder them in place. Note that all components are mounted vertically on the panel, and the layout is fairly cramped; insulated sleeving should be used where there is any danger of short-circuits occurring.

Before attempting to secure VRI and VR2 in place on the panel, the width of their mounting legs should be reduced, with the aid of a small file, so that they fit easily in the small holes in the Veroboard panel. Heat shunts should be used when soldering all semiconductors in place.

When assembly is complete, double check all wiring and ensure that no short-circuits are occurring between the copper strips on the underside of the panel. If satisfactory, the circuit should now be given a functional check and adjusted to give the correct output frequencies, as follows.
Temporarily wire VR3 and S1 to the unit, to conform to the circuit diagram, and connect the crystal and the battery in place. Switch on, and check on an oscilloscope that the \(100 \mathrm{kc} / \mathrm{s}\) signal is available at Tr 2 emitter. Now monitor the waveform at Tr 3 base; it should be possible to clearly see the sync pulses superimposed on the multivibrator waveform. Carefully adjust VRI until ten sync pulses are obtained for each complete cycle of the multivibrator; the unit is now operating at \(10 \mathrm{kc} / \mathrm{s}\).

Now monitor the voltage at Tr6 base, and adjust VR2 until ten sync pulses are again obtained for each complete cycle of the multivibrator; a \(1: 1\) mark/space ratio should be obtained. Check that \(\operatorname{Tr} 3\) is still operating correctly.

Fig. 3a (top left): Drilling and bending details of the "main" chassis.

Fig. 3b (centre left): Details for making the front panel.
Fig. 3c (below): Bending and drilling details for the crystal mounting and battery brackets.


Having completed the initial setting up, the rest of the unit can now be made up. Cut the main chassis, from light gauge aluminium, as shown in Fig. 3a. Next, make up the front panel, using a medium gauge aluminium, as shown in Fig. 3b. When ready the front panel should be covered with Fablon or a similar selfadhesive decorative plastic material; a material with a light woodgrained finish was used on the prototype.

Now make up the crystal holder bracket and the two battery holder brackets, as shown in Fig. 3c, and assemble them, as indicated in the inset, on the main chassis. Now secure the front panel to the chassis. Secure the Veroboard panel to the chassis using two 6BA screws passed through the mounting holes that are provided, and using two small rubber grommets interposed between the panel and the chassis to act as spacers! insulators. Fix S1, VR3, and Sk1 in place on the front panel, and complete the wiring up of the unit. When ready, the operating frequencies of the multivibrator sections of the unit should again be checked, as already outlined, and, if necessary, final adjustments should be made.

An attractive cabinet can be made up, with very little skill, as shown in Fig. 4. The two side pieces, the top, the base, and the rear panel should be cut from \(\frac{1}{8}\) in. hardboard, using the dimensions shown, and the four corner pieces should be cut from \(\frac{5}{8} \times \frac{7}{8}\) in. timber, and the unit should then be assembled as indicated. The hardboard parts should be nailed or screwed to the timber corner pieces, and care should be taken to ensure that all nails and screws are sunk flush with the outer surfaces of the cabinet. When assembly is complete, remove any rough spots with sandpaper, and then cover the entire cabinet with Fablon or a similar material.

The unit is held in place in its cabinet by means of four self-tapping screws passed through the holes at



Fig. 4: Dimensions and details for making a suitable case.
the corners of the front panel and screwed into the timber corner pieces of the cabinet. The unit can be given a final attractive finish by marking the front panel with pressure sensitive lettering, as indicated in the photographs.

The unit is now complete and ready for use; it should be noted, however, that fairly frequent checks should be made on the accuracy of frequency division of the multivibrator circuits throughout the life of the instrument.

\section*{\(\star\) components list}


\title{
practically wireless commentary by ILEITI
}

LONG as Henry may be in the tooth, decrepit as he may appear on one of those mornings after the gentlemen of the Press have been regaled by some self-seeking manufacturer, bloodshot as those orbs may seem, he can nevertheless hit a barn door with a well-aimed clod when called upon to do so.

Which makes it all the more annoying when a circuit has to be scanned with the sort of scrutiny your wife affords a post-party handkerchief, to find, not smudges of red, or any other colour, but conventional shapes, lines and symbols that custom has taught us ought to be in a particular place.

Custom may stale, and perhaps it is the same incentive our avantgarde artists have to present us with the unexpected which makes circuit drafters perform their convolutions. Whatever the reason, their efforts have an infinite variety. Searching for a particular component in some drawings is worse than digging for an errant sixpence under the eagle eye of a traffic warden.

Readers who become accustomed to the circuits that bespangle these pages may consider themselves well served. From jottings that would not disgrace the back of one of Alan Blumlein's envelopes, the careful lads of Newnes turn out an understandable print. The joints dot in where

. . . the eagle eye of a traffic warden.
they should, the cross-overs follow a set pattern, there is a satisfying constancy about the illustration of components. This is, in short, what is known as the "house style".

In the radio trade, a very different situation obtains. No two manufacturers think alike. In fact, one would be tempted to conjecture that many of them are as violently opposed to their rivals as they can be-even to the extent of wilfully making their service manuals just that little bit different.

Life would be dull if we were all the same. But life would be a lot easier if, in the technical manual field, at least, there was some measure of conformity. Some circuit diagrams defy analysis until one has taken a course in identification; where resistors and capacitors both appear as little boxes, where strange dots and triangles tell us what sort of component the ambiguous block is meant to be, provided we can find the key tucked away in the corner that always gets teastained first.

Provided, also, that we happen to have the visual acuity of a halfstarved hawk. Why do some illustrators specialise in the sort of lettering more used to scribing texts on the head of a pin than underlining vital items on a circuit diagram? And why are some of the circuits that have more parallel lines than the approach to Clapham Junction printed in a scale so minute that magnifying glass and miniature pointer must be used to trace an inter-connection? Is it the work of those devils, conspiring in their chapels, chuckling in their Gutenberg ale?

Some sadistic draughtsmen lay out their originals on one of those drawing boards the size of a hoarding that one sees in fascinating glimpses as one flashes past in the train. But, in an effort to eliminate the masses of close-

. . . just recovered from the dog. woven lines, some circuit designers bend over the other way. Here we find an output from stage \(B\) with an angled arrow marked 22. Somewhere on another part of the circuit there will be an answering arrow. By the time you have found it in the tangled mass of transistors, the set will have gone up in smoke.

On other types of circuit, there is no common earth or chassis return line. Each return is a dinky little black bar, easily overlooked and often unexpectedly "high" in the circuit layout-a dead trap for we oldies who still think that the voltage rises the higher up the drawing we go. The catch with these circuits is the odd audio circumstance when a true earth and a chassis return are not the same thing, or where some returns go to a common point to avoid the bugbear of hum loops.

Crossovers are a very dodgy business. There are four main systems: the hump-backed bridge where two wires cross but do not connect; the broken line that is almost the same thing, but which can be ambiguous; the plain line, except where dotted to show a joint, and the plain line which crosses except where it terminates against another line-no dots needed. There is a good case to be made out for all, but I fear those who conjure with the arguments do not have to crouch behind a radiogram in a dark corner of the lounge, with a crumpled wad of circuit just recovered from the dog.


1THIS tuner is primarily designed for plugging into the "Pyramid" amplifier, for immediate reception over the medium wave band. Long wave coverage can be added later, if wanted. The tuner could be used with other amplifiers, or for excellent phone reception, for personal listening.

Figure 1 is the circuit, and current is drawn from the amplifier. L1 is a ferrite rod winding, dispensing with any need for an external aerial. Ll and the oscillator coil L2 are tuned by the ganged capacitor \(\mathrm{VC1} / \mathrm{VC2}\), which has trimmers TCl and TC 2 , and an oscillator padder C3.

There are two stages of intermediate frequency amplification, followed by the diode D1, which provides audio signals and automatic volume control bias through R12 to Tr 2 . The amplifier volume control supplies the diode load. If the tuner is used with other equipment, a \(5.6 \mathrm{k} \Omega\) resistor may need wiring across C11, for this purpose. Medium or high impedance phones may be connected across CII. A straightforward circuit of this type is generally easy to build and align. When used in conjunction with the "Pyramid" amplifier, a large number of stations can be received at excellent volume.


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\hline 1 AF & \(5 /\) & \(6 \mathrm{~F} 6 \mathrm{M} \quad 7 / 8\) & 12BA6 \(5 /-\) & 185BT 85/- & DY87 \(5 / 9\) & EL35 10/- & P61 2/6 & TP2620 8/9 & \(\times 41\) 10\% & BCl15 3/- & OA202 2j- \\
\hline 1A7GT
1 Cl & \(7 / 6\)
\(4 / 8\) & \(\begin{array}{ll}\text { 6F12 } & 3 / 8 \\ 6 \mathrm{Fl13} & 3 / 8\end{array}\) & \(\begin{array}{ll}\text { 12BE6 } & 5 / 3 \\ 12 \mathrm{BH7} & 6 / 6\end{array}\) & 301 20/- & E80F 84/- & EL36 \(8 / 9\) & PABC80 \(7 / 6\) & TY86F 1212 & \(\times 61\) 8/9 & BC116 5/- & OA210 9/6 \\
\hline \({ }_{1 \mathrm{Cl}}^{1 \mathrm{C}}\) & \(4 / 9\)
\(7 / 9\) & \(\begin{array}{cr}6 \mathrm{~F} 13 & 3 / 8 \\ 6 \mathrm{~F} 14 & 15 \%\end{array}\) & 12EH7
12 E
17/6 & \(\begin{array}{ll}302 & 16 / 6 \\ 303 & 15 / 6\end{array}\) & \(\begin{array}{ll}\text { E83F } & \text { 24/- } \\ \text { E88CC } & 18 /-\end{array}\) & \(\begin{array}{cr}\text { EL37 } & 16 / 6 \\ \text { EL41 } & 9 / 3\end{array}\) & PC86
PC88
P10 & UABC80 5/9 & X64 & BC118 \(4 / 6\) & OA211 13/6 \\
\hline 1 C 3 & \(6 / 6\) & \(6 \mathrm{FW} 1610 / 9\) &  & \(\begin{array}{ll}303 & 15 /- \\ 305 & 16 / 6\end{array}\) & \(\begin{array}{lll}\text { E88CC } & 18 /- \\ \text { E180F } & 17 / 6\end{array}\) & \(\begin{array}{ll}\text { EL41 } & 9 / 3 \\ \text { EL42 } & 9 /\end{array}\) & \(\begin{array}{ll}\text { PC88 } & 9 / 9 \\ \text { PC95 } & 8 / 8\end{array}\) & \(\begin{array}{ll}\text { UAF42 } & 9 / 6 \\ \text { UB41 } & 10 / 6\end{array}\) & \begin{tabular}{l}
\(\times 65\) \\
\(\times 66\) \\
\(\times 66\) \\
\hline
\end{tabular} & BDI19 9/- & OAZ20012/- \\
\hline 1 C 5 & \(4 / 9\) & \(6 F 17\) 12/6 & 12J7GT \(6 / 6\) & \(\begin{array}{ll}305 & 16 / 6 \\ 306 & 13 /=\end{array}\) & EA50 \(1 / 6\) & \(\begin{array}{ll}\text { EL42 } & 9 / 0 \\ \text { EL81 } & 8 /-\end{array}\) & \(\begin{array}{ll}\text { PC95 } & 8 / 8 \\ \mathrm{PC97} & 7 / 8\end{array}\) & \(\begin{array}{cc}\text { UB41 } & 10 / 6 \\ \text { UBC41 } \\ 7 / 8\end{array}\) & \begin{tabular}{ll}
\(\times 66\) & \(7 / 6\) \\
\(\times 76 \mathrm{M}\) \\
\hline
\end{tabular} & BFY50 5/- & OAZ \(20110 / 6\) \\
\hline \(1 \mathrm{C6}\) & \(10 / 6\) & 6 F 18 8/8 & 12K5 11/6 & 807 11/9 & EA76 18\% & EL83 8/9 & PC900 \(9 /\). & \begin{tabular}{ll} 
UBC4 \\
UBC81 & \(7 / 8\) \\
\hline 186
\end{tabular} & \(\times 76 \mathrm{M}\) 7/9 & BFY51 4/6 & OAZ202 9/- \\
\hline 119 & \(6 / 8\) & \(61^{4} 1312 / 3\) & \(12 \mathrm{K7GT} 3 / 6\) & 856 & EABC80 8/- & EL84 4/6 & PCC84 8/- & \(\begin{array}{ll}\text { UBC81 } & 6 / 6 \\ \text { UBF80 } & 5 / 9\end{array}\) & X81M \(29 / 1\) & BFY52 5/- & OAZ203 9/6 \\
\hline 106 & 9/6 & \(6 \mathrm{~F}^{2} 4\) 10\% & 12K8GT \%/8 & 1821 10/0 & EAC91 3/8 & EL85 7/6 & PCC85 6/9 & UBF80 \(5 / 8\) & \(\times 101\) 2971 & BFI54 5/- & OAZ204 \(01-\) \\
\hline 1FD1 & 61- & 6 F 25 10/- & 12Q7GT 3/6 & 5763 10/- & EAF42 8/3 & EL86 8/- & PCC88 11/- &  & X109 28/- & BF'159 5/- & OAZ205 9/- \\
\hline 1FD9 & 3/9 & 6F28 10/6 & 12SA7GT818 & 7193 10/8 & EB34 7/6 & EL91 2/6 & PCC89 10/3 & UC92 5/6 & Y65 5\% & \(\begin{array}{ll}\text { BF163 } \\ \text { BF167 } & \text { 2/- } \\ \text { 2/6 }\end{array}\) & OAZ206 9/- \\
\hline 166 & 8/- & 61732 8/- & \(12 \mathrm{SC7}\) 4/- & 7475 4/. & EB41 4/9 & EL95 5/- & PCC189 9/3 & UCC84 8/- & Z63 4/9 & \({ }_{\text {BF173 }}\) & OAZ40710/6 \\
\hline 1H5GT & 71- & 6G6G \(2 / 6\) & 12847 8/- & A1834 20/- & EB91 \(2 / 3\) & EM71 14/- & PCF80 616 & UCC85 6/6 & 277 3/3 & BF180 12\% & OAZ210 7/- \\
\hline 11.4 & 2/6 & 6H6GT 1/6 & 128.57 6/- & ACO44 14/- & EBC3 20/6 & EM80 5/9 & PCF82 6/3 & UCF80 8/8 & Z329 11/9 & BF185 81- & OAZ224 \\
\hline \(1 \mathrm{LD5}\) & \(5 /-\) & \(6 J 5 G \quad 3 / 9\) & \(12 \mathrm{SK} 78 /-\) & AC2PEN & EBC41 9/6 & EM81 7/6 & PCF84 8\% & UCE21 9/- & Z729 6/3 & BTX34/400 & OAZ224 \({ }^{16 / 6}\) \\
\hline 1LN5 & \(4 / 6\) & 6JJGT 4/6 & 12897aT8/- & 19/8 & EBC81 6/3 & EM84 6/- & PCF86 8/9 & UCH42 10/- & Z759 28/ & 40/0 & OC19 25/- \\
\hline 1N5GT & 718 & 6J6 6 \% & 128 F 7 5/- & AC2PEN/ & EBC90 4/- & EMB5 11\% & PCF801 7/- & UCH81 6/\% & 375 & BY100 \({ }^{\text {a/8 }}\) & \(0 \mathrm{C22}\) 5/- \\
\hline 1 Pl & \(7 / 8\) & 6J7G 4/9 & \(12 Y 4\) 2/- & DD 18/6 & EBC91 5/- & EM87 7/3 & PCF802 \(9 / 6\) & UCL82 7/- & Transistors & BY101 11/6 & 0 C 23 7/- \\
\hline 1 P 10 & \(4 / 9\) & 6J7GT 8/6 & 13 Dl 5/- & AC6PEN 4/8 & EBF80 5/9 & EY51 0/6 & PCF805 818 & UGL83 \(9 / 6\) & and diodes & BY105 10/6 & OC24 \(14 / 6\) \\
\hline 1P11 & \(5 / 6\) & \({ }_{6 K 6 G T} 5 /-\) & 13D3 \(9 \%\) & AC/PEN (5) & EBF83 7/- & EY81 7/- & PCF80611/6 & UF41 \(9 / 8\) & 2G225 10/6 & BY114 6/6 & 0025 5/- \\
\hline 1R5 & \(4 / 9\) &  & \(\begin{array}{lr}14 \mathrm{H7} & \text { 9/6 } \\ 1497 & 19 / 8\end{array}\) & 19/6 & EBF89 \(8 / 9\) & EY83 8/3 & PCF80812/6 & UF42 9/- & 2 N 404 6/- & BYL20 6/6 & 0 Ca 2 5 5- \\
\hline 184 & \(4 / 9\) & \begin{tabular}{l}
6 KFGT \\
6 K 8 Cl \\
\hline \(1 / 6\)
\end{tabular} & \(1497 \quad 19 / 8\) & EN (7) & EBL21 \(10 / 8\) & EY84 9/6 & PCL81 9/- & UF80 6/9 & 2N2297 4/6 & BY234 4/7 & OC28 5/, \\
\hline 185 & \(3 / 9\) & 6K8G 6 K 8 GT 7/8 & \(\begin{array}{ll}18 & 12 / 6 \\ 19 & 10 / 6\end{array}\) & 19/6 & EC52 4 [ 413 & EY86 6/- & PCL82 616 & UF85 713 & 2N2369A4/3 & BY236 4/- & OC29 16/6 \\
\hline 172 & 34/11 & \(\begin{array}{ll}\text { 6K8GT } & 7 / 8 \\ \text { 6L6GT } & 7 / 9\end{array}\) & \(\begin{array}{lr}19 & 10 / 6 \\ 19 \text { AQ5 } & 4 / 8\end{array}\) & AC/THI 10/- & EC53 12/6 & EY87 8/- & POL83 8/9 & UF86 9/- & 2N31+21 50/- & BY238 4/- & OC30 7/- \\
\hline \(1 T 4\)
104 & \(2 / 9\)
\(5 / 8\) & \(\begin{array}{lr}\text { 6L6GT } & 7 / 9 \\ 6 \mathrm{~L} 1 & 19 / 6\end{array}\) & \begin{tabular}{ll}
\(19 \mathrm{AQ5}\) & \(4 / 9\) \\
19 El & \(40 \%\) \\
\hline 18
\end{tabular} & AC/TP 19/6 & \(\begin{array}{ll}\text { EC54 } & 6 /- \\ \text { EC70 } & 4 / 8\end{array}\) & EY88 7/6 & PCL84 713 & UF89 5/9 & 2N3866 20j- & BYY Y 3 \(30 /-\) & \(0 \mathrm{O} 5510 \%\) \\
\hline 1 LU & \begin{tabular}{l} 
5/6 \\
\(5 / 3\) \\
\hline 18
\end{tabular} & \(\begin{array}{lr}6 \mathrm{~L} 18 & 19 / 6 \\ 6 \mathrm{~L} 18 & 7 / 6\end{array}\) & \(\begin{array}{ll}19 \mathrm{El} & 40 \% \\ 201) \\ 13 \%\end{array}\) & AC/VP112/. & \(\begin{array}{lr}\text { EC70 } & 4 / 9 \\ \text { EC86 } & 10 / 8\end{array}\) & EY91 3/* & \(\begin{array}{ll}\text { PCLA5 } & 8 / 3 \\ \mathrm{PCL86} & 8 / 3\end{array}\) & ULA1 9/- & AA120 3/- & BYZ10 5/- & 0C36 7/6 \\
\hline 2 A 7 & 12/6 & 6 L 19 18\% & 20D4 20/5 & AC/VP2 11/- & C88 \(10 / 8\) & EZ35 5/- & 86 8/3 & UL46 9/6 & AA129 3/- & BYZ11 5/- & OC38 11/6 \\
\hline 2 D 13 C & 7/- & \(6 \mathrm{LD} 20 \quad 6 / 6\) & \(20 \mathrm{~F} 214 / \mathrm{F}\) & \(\begin{array}{ll}\text { ATP4 } & 8 / 3 \\ A^{2} & 81\end{array}\) & EC92 8/6 & EZ40 7/3 & 15/= & UL84 6/- & AAZ13 3/6 & BYZ12 50- & OC41 10/ \\
\hline 2D21 & 8/6 & 6N7GT 7/- & 20 L 1313 & \begin{tabular}{ll} 
AZ1 \\
\hline 1891 \\
816
\end{tabular} & ECC31 15/6 & EZ41 7/3 & \(45{ }^{7 /}\) & \(51-\) & 107 3/6 & BYZ13 5\%- & OC42 6/9 \\
\hline \(2 \times 2\) & 3/- & \(6 \mathrm{P}^{1} 1\) 12/- & 20P1 17/6 & \(\begin{array}{ll}\text { AZ31 } & 8 / 9 \\ A Z 41 & 8 / 8\end{array}\) & ECO32 4/6 & E280 3/9 & 19/6 & \(\begin{array}{ll}\text { UR1C } & 10 / 6 \\ \text { UU5 } & 7 /-\end{array}\) & AC119
AC114
\(8 /-\) & BYZ15 35/- & \({ }_{0}^{0 C 43} 18 / 6\) \\
\hline 3 A 4 & 818 & 6 P 25 12/- & \(20 \mathrm{P3}\) 18/- & \begin{tabular}{ll} 
AZ41 \\
B36 & \(8 / 8\) \\
\hline 19
\end{tabular} & ECC33 89/1 & \(\begin{array}{ll}\text { EZ81 } & 4 / 6 \\ \text { E790 } & 3 / 6\end{array}\) & PEN46 4/- & UUS 16/6 & \(\begin{array}{ll}\text { AC114 } \\ \text { AC126 } & \text { 8/- }\end{array}\) & \(\begin{array}{ll}\text { CG12E } & 4 /- \\ \text { CG64H } & 4 / \mathrm{c}\end{array}\) & OC44 0 2/- \\
\hline 3A5 & \(8 / 6\) & \({ }^{6 P 26} 1818 /-\) & \(20 \mathrm{P4} 178\) & B319 8/- & ECC34 29/6 & LW4/5006/6 & PEN383 9/6 & UU12 \(4 / 6\) & ACl27 2/- &  & \(0 \mathrm{C45} 1 / 9\) \\
\hline 3 B 7 & 5/- & \(\begin{array}{ll}6 \mathrm{Pr28} & 25 /- \\ 607 \mathrm{~g} & 5 /-\end{array}\) & \(20 \mathrm{P5}\) 17/\% & \({ }_{\text {BL63 }} 10 / 6\) & ECC35 \(4 / 9\) & FW4/8008/6 & PEN384 & UY1N 10/3 & ACl28 2j- & GD4 \(6 / 6\) & OC45M 8/- \\
\hline \(3 \mathrm{B6}\) & \(8 / 8\)
\(5 / 8\) & \(\begin{array}{ll}\text { 6Q7G } & 5 /- \\ 6079 \mathrm{~T} & 8 / 9\end{array}\) & \({ }_{25}^{25 A G G} \quad 7 / \mathrm{B}\) & CK506 \(0 / 6\) & ECC40 \(9 / 6\) & FW4/8008/6 & 11/6 & UY21 9/- & ACL54 5/- & G115 \(\quad\) b/6 & 0 O 46 3/- \\
\hline \(3 \mathrm{C4}\) & \(5 / 3\)
\(6 / 6\) & \(\begin{array}{ll}\text { 6Q7GT } & 8 / 9 \\ \text { 6R7G } & 5 / 6\end{array}\) & \(\begin{array}{ll}25 L 6 \mathrm{Ca} & 8 / 9 \\ 25 \% 5 & 6 /-\end{array}\) & \(\begin{array}{ll}\text { CL4 } & 19 / 6\end{array}\) & ECC81 \(4 /-\) & G230
\(\mathbf{G Z 3 2}\)
\(9 /-\) & PEN453DD & UY41 \(6 / 6\) & AC155 \(6 / 6\) & GD6 5/6 & OC65 22/6 \\
\hline 3 BESGT & \(6 / 6\)
\(4 / 8\) & 6R7G
6847 T
\(7 / 6\) & \(\begin{array}{ll}25 Y 5 & 6 /- \\ 25 Y 59 & 8 / 6\end{array}\) & CL33 19/6 & \(\begin{array}{ll}\text { ECC82 } & 4 / 6 \\ \text { ECC83 } & 4 / 6\end{array}\) & \(\begin{array}{ll}\text { G233 } & \text { 12/6 } \\ \text { G23 } & 18\end{array}\) & 19/6 & UY85 5/8 & ACl56 4]- & GD8 4/- & 0 OCO 2/3 \\
\hline 384
3 V 4 & \(4 / 8\)
\(5 / 6\) & 68A7GT \(6 / 6\) & \(\begin{array}{ll}25 Y 5 G & 8 / 6 \\ 25 \mathrm{Z4G} & 6 / 8\end{array}\) & CV6 \(10 / 6\) & \(\begin{array}{ll}\text { ECC83 } & 4 / 6 \\ \mathrm{ECC84} & 8 /-\end{array}\) & \(\begin{array}{ll}\text { Q234 } & 101- \\ & \text { 12/ }\end{array}\) & PENA4 19/6 & \(\begin{array}{ll}\text { U10 } & 8 /- \\ \text { U12 } & 7 / 6\end{array}\) & AC157 5/- & GD9 4/- & 0 OCl 2/- \\
\hline 4D1 & \(8 / 9\) & 68G7 719 & \(25 \mathrm{Z5}\) \%/. & CV63 10/6 & ECC85 5\% & \({ }_{\text {GZ37 }}\) 14/8 & PEN/DD \({ }_{\text {4020 }} 17 / 6\) & \(\begin{array}{lll}\text { U12/14 } & 7 / 6 \\ \text { U15 }\end{array}\) & AC165
AC166
5/- & QD10 4/- & \(0 \mathrm{C72}\) 21- \\
\hline 5 F 4 GY & \(8 / 9\) & \(68 \mathrm{H7} 3 /-\) & 25269 8/6 & CV271 12/6 & ECC88 & H30 5/- & 4020
PFL20012/6 & \(\begin{array}{ll}\mathrm{U} 16 & 15 /- \\ \mathrm{U} 17 & 5 / 7\end{array}\) & AC166
AC167
12\% & GD11 4/- & \(0 \mathrm{OC73} 181-\) \\
\hline SU4G & 418 & 68 J 7 5/- & \(30 \mathrm{C1} \quad 6 / 6\) & CV428 18/- & ECO189 9/6 & HABC80 \(9 / 8\) & PFL20012/6 & \(\begin{array}{ll}\text { U17 } & 5 / 7 \\ \text { U18/20 } & 6 / 6\end{array}\) & \(\begin{array}{ll}\text { AC167 } \\ \text { ACl } 68 & 12 / 6\end{array}\) & GD12 \({ }_{\text {GD14 }}\) 4/- & \(\begin{array}{ll}0074 \\ 0 C 75 & 8 /- \\ 0\end{array}\) \\
\hline 6V4G & 8/- & 68K7 4/6 & 30015 13/6 & CYI 16/4 & ECC80412/6 & HL2 7/6 & PL33 9/- & U19 40/- & AC169 6/6 & \begin{tabular}{ll} 
GD14 \\
GD15 & \(8 \%\) \\
\hline
\end{tabular} & \begin{tabular}{ll}
\(0 C 75\) \\
0076 & \(2 /-\) \\
\hline
\end{tabular} \\
\hline 5Y3GT & \(5 / 9\) & 68L7GT 4/9 & 30017 13/- & CY1C 10/6 & ECC807 27/- & HL13C 4/- & \(\begin{array}{lr}\text { Pl3 } & 19 / 9\end{array}\) & U22 6/9 & AC176 11/. & GD16 4/- & \(\begin{array}{ll}0076 & 3 /- \\ 0077 & 3 / 4\end{array}\) \\
\hline 6\%3 & \(7 / 6\) & 68N7GT 4/6 & 30018 8/9 & CY31 7/9 & ECF80 7\%- & HL23DD 5/- & \({ }_{\text {PL81 }}\) & U25 13/- & AC177 5/6 & GET102 4/9 & \(\begin{array}{ll}0077 & 3 / 4 \\ 0078 & 3 / 4\end{array}\) \\
\hline 5Z4G & 7/6 & 68Q7GT 61- & \(30 \mathrm{~F} 511 / 8\) & D1 1/8 & ECF82 8/9 & \(\mathrm{HLA1}^{\text {8/9 }}\) & PL81A 7/6 & U26 10/6 & ACY17 3/4 & GET103 4/- & \(\begin{array}{ll}0 C 78 & 3 /- \\ 0078 \mathrm{D} & 3 /-\end{array}\) \\
\hline 6/30L2 & \(12 / 6\) & \({ }^{6887}\) \%/- & \(30 \mathrm{FL1}\) 15/- & D15 15/6 & ECF86 9/- & HLalod \({ }^{\text {de/6 }}\) & PL82 5/9 & U31 8/3 & ACYt8 \(5 / 8\) & GET10518/- & OC79 \\
\hline 6 A 8 G & 8/8 & 6U4GT 9/6 & \(30 \mathrm{FLl2} 15 /-\) & D63 5/- & ECF804 42/- & \(18 / 6\) & PL83 6/- & U33 13/6 & ACY19 6/8 & GET111 & \(0 \mathrm{C81} \mathrm{2/-}\) \\
\hline 6AC7 & 81- & \({ }^{645 G} 5 /-\) & \(30 \mathrm{FL} 1412 / 6\) & \({ }^{\text {D77 }}\) 2/3 & ECF80512/6 & HL42 DD 8/- & \({ }^{\text {PL84 }} 818\) & U35 16/6 & ACY20 4/9 & 15/6 & OC81D \(2 /-\) \\
\hline 6AG5 & \(2 / 6\) & 6U7G 7/- & 30 Ll 6/- & DAC32 7/- & ECH21 9/6 & HN309 27/4 & PL302 11/- & U37 34/11 & ACY'21 5/9 & GET113 4/- & OC81M \({ }_{\text {B }}\) \\
\hline 6AG7 & 5/8 & \({ }_{6}^{6 V 6 G}\) & \(30 \mathrm{L15}\) 18/9 & DAF91 3/8 & ECH33 \(22 / 8\) & HVR2 8/8 & PL500 12/8 & U45 1516 & ACY22 3/6 & GET11517/- & \(0 \mathrm{C812} 5\) \\
\hline 6AJ5 & \(8 / 6\) & \({ }^{6 \times 6 G T}\) 6/6 & \(30 \mathrm{L17}\) 13/- & DAF96 \(8 / \mathrm{l}\) & ECH35 5/9 & HVR2A 8/8 & PL504 13/- & U50 5/9 & ACY28 4/3 & GET116 7/6 & per pair \\
\hline 6AK5 & 4/8 & \(\begin{array}{lll}6 \times 4 & 8 / 6 \\ 6 \times 5 \mathrm{GT} & 5 / 8\end{array}\) & \(30 \mathrm{P4} 12 /-\) & \(\begin{array}{ll}\text { DCC90 } & 8 / 8 \\ \text { DD4 } & 10 / 8\end{array}\) & ECH42 9/6 & IW3 \(5 / 8\) & PL802 15/- & U52 4/9 & AD140 8/- & GET119 4/6 & \(0 \mathrm{C82}\) 2/3 \\
\hline 6AK6 & 61. & \(\begin{array}{ll}\text { 6X5GT } \\ \text { 6Y7G } & \text { 12/6 }\end{array}\) & 30P4MR & \(\begin{array}{ll}\text { DD4 } & 10 / 8 \\ \text { DD41 } & 12 / 6\end{array}\) & ECH81 \(6 / 6\) & IW4/350 5/6 & PM54 9/8 & U76 4/9 & AD149 8/- & GET573 8/6 & OC82D \(2 / 6\) \\
\hline 6AK8 & 6/- & \(\begin{array}{ll}6 Y 7 G & 12 / 6 \\ 747 & 12 / 6\end{array}\) & \({ }_{30 \mathrm{P} 12}{ }^{17 / 6}\) & \(\begin{array}{lr}\text { DD4 } & 12 / 8 \\ \text { DDT4 } & 7 / 6\end{array}\) & ECR83 7/9 & TW4/500 6/E & PX4 14/- & U78 8/6 & AF'102 18/- & GET5878/6 & 0 O 83 2/- \\
\hline 6AL5 & \(2 / 8\)
\(16 / 6\) & \(\begin{array}{lr}\text { 7A7 } & \text { 12/6 } \\ \text { 7AN7 } & 6 /-\end{array}\) & \(\begin{array}{ll}30 \mathrm{Pl2} & 13 /- \\ 30 \mathrm{P} 19 & 11 /-\end{array}\) & \(\begin{array}{ll}\text { DDT4 } & 7 / 8 \\ \text { DF33 } & 7 / 9\end{array}\) & \(\begin{array}{ll}\text { ECH84 } \\ \text { ECL80 } & 7 /= \\ \end{array}\) & KBC32 \({ }_{\text {KF35 }}^{\text {K }}\) 12/5 & \(\begin{array}{ll}\text { PY31 } & 6 / 6 \\ \text { PY } 32 & 9 / 8\end{array}\) & \(\begin{array}{ll}\mathrm{Ul07} & 17 / 6 \\ \mathrm{U191} & 12 / 6\end{array}\) & AF114 4/- & GET87210/- & OCB4 3/- \\
\hline 6AM5 & 2/6 & 7B6 10/9 & \(30 \mathrm{PL1}\) 15/- & DF72 30/- & ECL82 6/6 & KL35 11/6 & PY32 9/6 & \(0191{ }^{12 / 6}\) & AF115 3/- & GET873 4/- & \(\mathrm{OCl}^{123}\) 4/6 \\
\hline 6AM6 & 8/8 & \({ }_{7} 7 \mathrm{B7}\) 7\%- & \(30 \mathrm{PL13} 15 /-\) & DF91 \(2 / 9\) & ECL83 9/- & KLL32 \(21 / 7\) & \(\begin{array}{ll}\text { PY33 } \\ \text { PY80 } & \text { 9/6 } \\ \text { 5/- }\end{array}\) & \(\begin{array}{lr}\text { U251 } & 12 / 6 \\ \mathrm{U} 281 & 8 / 9\end{array}\) & \(\begin{array}{ll}\text { AFl16 } \\ \text { AFl17 } & 3 / 4\end{array}\) & 23/8 & \(0 \mathrm{OL139}\) 12/- \\
\hline 6AQ5 & \(4 / 9\) & 7 CD 40\% & 30PL14 15/- & DF96 6/- & ECL84 12/- & KT2 5/- & PY881 5/\% & \begin{tabular}{lr} 
U281 & 12/8 \\
\hline 2818
\end{tabular} & AF117
AF119
8/4 & GET88210/6 & \(\begin{array}{ll}006140 & 19 /- \\ 00169 & \\ \\ 0\end{array}\) \\
\hline BAR6 & 201- & 768 6/6 & \(30 \mathrm{PL15} 15 /-\) & DF97 10/- & ECL85 11/- & KT8 15/- & PY82 5/- & \(\begin{array}{ll}\text { U301 } & 18 / 6\end{array}\) & AF124 7/6 & GE1'887 4/6 & 0C170 2/6 \\
\hline 6AT6 & 4/- & 7H7 5/6 & \(35 \mathrm{A5}\) 15/- & DH30 15/6 & ECL86 719 & KT32 4/9 & PY83 \(\quad 5 / 6\) & U329 12/6 & AF125 3/6 & GET889 4/6 & \(0 \mathrm{Cl71}\) 3/4 \\
\hline 6AU6 & 5/6 & 7R7 12/6 & \(35 \mathrm{DF} 11 / 9\) & DH63 5/- & ECLL800 & KT41 19/6 & PY88 6/8 & \(\begin{array}{ll}\mathrm{U} 403 & 6 / 6\end{array}\) & AF126 7/- & GET890 4/6 & \(0 \mathrm{Cl72}\) 4/- \\
\hline 6AV6 & 51. & 7V7 5/- & 35L6GT 6/8 & DH76 8/6 & 301- & KT44 5/8 & PY800 6/8 & U404 7/6 & AF'127 3/6 & GET896 4/6 & 0 C 200 5/. \\
\hline \(6 \mathrm{B8G}\) & 2/6 & \(7 \mathrm{Y} 4 \quad 6 / 6\) & \(35 W 4 \quad 4 / 6\) & DH77 4/- & EF22 12/6 & KTb1 12/- & PY801 8/6 & U801 18/- & AF139 11/- & GET897 \(4 / 6\) & \(0 \mathrm{C201}\) 28/- \\
\hline \(6 \mathrm{BA6}\) & 4/6 & 9BW6 \(\quad 9 / 6\) & \(35 \mathrm{Z8}\) 10/- & DH81 10/9 & EF36 8/- & KT63 4/- & P230 9/6 & U4020 6/0 & AF178 10/- & GEX13 8/6 & 00202 \\
\hline \(6 \mathrm{BE6}\) & \(4 / 8\) & 9D7 716 & \(35 \mathrm{Z4GT} 4 / 9\) & DH101 25/- & EF37A 7/- & KT66 16/6 & QP21 5/- & VP4B 11/- & AFl79 13/6 & GEX35 4/6 & OC203 5/6 \\
\hline 6BG6G & \(20 / 5\) & 10C1 \(12 / 6\) & \(35 \mathrm{Z5GT} 5 / 6\) & DH107 & EF39 5/- & \(\begin{array}{lr}\text { KT74 } & 18 / 6 \\ \text { KT76 } & \\ 7 / 6\end{array}\) & QQV03/10 & VP13C 7/- & AF180 9/6 & GEX \(3610 /-\) & \(0020410 / 6\) \\
\hline \({ }_{6}^{68 \mathrm{BH} 6}\) & 7/- & \(\begin{array}{ll}10 \mathrm{C} 2 & 18 /- \\ 10 \mathrm{Dl} & 7 / \%\end{array}\) & \(\begin{array}{lr}42 & 5 /- \\ 43 & 10 /-\end{array}\) & DK32 16/11 & EF40
EF41
8/9 & \(\begin{array}{lr}\text { KT76 } & 7 / 6 \\ \text { KT88 } & 29 / 6\end{array}\) & 8875/20/- & VP23 2/6 & AF181 14/- & GEX45/17- & \(00^{0205}\) 7/6 \\
\hline 6BJ6
6 BQS & \(8 / 9\)
\(4 / 6\) & \(\begin{array}{ll}10 \mathrm{D} 1 & 7 / 7 \\ 10 \mathrm{D} 2 & 14 / 7\end{array}\) & \(\begin{array}{ll}43 & 10 /- \\ 5045 & 21 / 10\end{array}\) & \(\begin{array}{lc}\text { DK32 } & 7 /- \\ \text { DK40 } & 10 / 6\end{array}\) & \(\begin{array}{cc}\text { EF41 } & 9 /- \\ \text { EF42 } & 8 / 6\end{array}\) & \(\begin{array}{llr}\text { KT88 } & \text { 28/6 } \\ \text { KTW61 } \\ \text { K/g }\end{array}\) & Q875/20 \({ }^{10 / 6}\) & VP41
VR75
V/- &  & GEXES/1 & 00612 8/- \\
\hline 6BQ7A & 71 & 10 Fl 15/- & 50B5 6/3 & DK91 4/9 & EF50 2/6 & KTW6218/6 & \(15^{10}\) & VR715 24/- & ASY27 8/6 & 15/* & OCP71 27/6 \\
\hline \(6 \mathrm{BR7}\) & 9/- & \(10 \mathrm{F9}\) 9/- & \(50 \mathrm{C5} 519\) & DK92 \(7 / 9\) & EN54 6/- & KTW63 6/- & Q8150/15 & VR159 5\% & A8Y28 \(6 / 6\) & GHX 66 16/- & ORP12 15/6 \\
\hline 6BR8 & 8/. & \(10 \mathrm{Fl} 88 / 6\) & \(50 \mathrm{CD6G441/}\) & DK96 6/6 & EF73 6/6 & KTZ41 6/- & R10 15/- & VT61A 7/- & AY100 \(88 \%\) & M1 2/10 & T32 12/8 \\
\hline \({ }^{6 B 87}\) & 18/6 & 10LD3 \(7 / 8\) & 50L6GT 6/- & DL33 6/6 & EF80 4/6 & LN309 8/9 & R11 19/6 & VT501 3/- & BA115 2/8 & M3 2/10 & T93 15/- \\
\hline \(6 \mathrm{BW7}\) & 5/6 & 10LD11 14/6 & 52 KU 14/6 & DL35 4/9 & EF83 \(9 / 9\) & LP2 9/6 & R12 616 & VU111 8/- & BA116 9/\% & OA5 5/6 & V10/15 \({ }^{\text {a }}\) \\
\hline 6 BXB & \(4 / 6\) & \(10 \mathrm{P} 1315 / 8\) & \({ }^{53 \mathrm{KUU}} 14 / 6\) & DL72 15/- & EF85 4/9 & LZ319 6/6 & H16 34/11 & VU120 12/- & BA129 2/6 & OA9 2/6 & 12/- \\
\hline \(6 \mathrm{C5GT}\) & \(8 /\) & 10 P 14 15/8 & \(72 \quad 6 / 6\) & DL75 30\% & EF86 6/8 & LZ329 6/8 & R17 17/6 & VU120A18/- & BA130 2/- & OA10 6/6 & X \(\triangle 102\) 19/6 \\
\hline 6 C 6 & \(8 / 8\) & 1246 6/- & 77 5/- & DL92 4/9 & EF89 4/9 & MHD4 \(7 / 6\) & \(\mathrm{R18} 916\) & VU133 7/- & BCY10 5j- & OA47 2/- & XA103 15/- \\
\hline 6 Cl 9 & \(12 / 6\) & 12AC6 0f- & 78 4/9 & DL94 5/6 & EF91 3/3 & MHLD6 & R19 6/6 & W42 11\% & BCY12 5/- & OA70 8/- & MAT1007/9 \\
\hline \({ }_{6}^{6 C D O}{ }^{\text {a }}\) & 19/6 & 12AD6 \(10 / 3\) & 85 A2 \(8 / 6\) & DL96 7/6 & EF92 216 & 12/6 & RK34 7/6 & W61M 24/6 & BCY 33 \$/- & 0 0A73 8/- & MAT1018/6 \\
\hline 6CD7 & 9/8 & 12AE6 8/6 & 90AG 67/6 & DLS10 10/6 & EF97 8\% & MU12/14 4/0 & SP13C \(12 / 6\) & W63 10/6 & BCY34 5/- & 0479 1/9 & MAT120 7/9 \\
\hline \({ }^{60186}\) & 6/- & 12A'T6 4/8 & 90AV 67/6 & DM70 7/6 & EF98 10/6 & MX \(4018 / 6\) & \(\mathrm{gP42}^{12 / 6}\) & W76 3/6 & BCY 38 5/\% & OA81 1/9 & MAT121 8/6 \\
\hline 6CW4 & 18\% & \(12 \mathrm{AT7} \mathrm{4/-}\) & 90CG 34/- & DM71 7/6 & EF183 6/3 & N37 88/8 & 8P61 2/m & W77 2/6 & BCY39 \%/- & OA85 1/6 & ZE12V7 1/9 \\
\hline
\end{tabular}

MATCHED TRANSISTOR BETS 1-OC44 and 2-OC45 8/6; 1-OC81D and 2-OC81 7/6;1 -0C82D and 2-0C82 8/6; Set of three-0C83 (OET118/119) 8/6; LP15 package (AC113, AC154, ACI57, AA120) 12/6; Postage 6 d . per set.

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}

\section*{components list}
\begin{tabular}{ll}
\multicolumn{2}{l}{ Resistors: } \\
R1 & \(56 \mathrm{k} \Omega\) \\
R2 & \(10 \mathrm{k} \Omega\) \\
R3 & \(3 \cdot 9 \mathrm{k} \Omega\) \\
R4 & \(680 \Omega\) \\
R5 & \(56 \mathrm{k} \Omega\) \\
R6 & \(680 \Omega\) \\
R7 & \(4 \cdot 7 \mathrm{k} \Omega\) \\
R8 & \(22 \mathrm{k} \Omega\) \\
R9 & \(1 \mathrm{k} \Omega\) \\
R10 & \(1 \cdot 2 \mathrm{k} \Omega\) \\
R11 & \(3 \cdot 9 \mathrm{k} \Omega\) \\
R12 & \(8 \cdot 2 \mathrm{k} \Omega\) \\
All & \(10 \% \frac{1}{4}\) watt
\end{tabular}

\section*{Semiconductors:}
\(\begin{array}{llll}\text { Tr1 } & \text { NKT152 or OC44 } & \text { Tr3 } & \text { NKT154 or OC45 } \\ \text { Tr2 } & \text { NKT153 or OC45 } & \text { D1 } & \text { OA70, OA81, etc. }\end{array}\)
Inductors:
L1 Medium wave ferrite rod aerial for OC44 or similar, Osmor QFR2B etc. for 208pF tuning.
L2 Oscillator coil for OC44 or similar, Osmor Red Spot etc., for 176 pF .
IFT1 and IFT2, 1 st and 2nd IFT's for OC45 or similar, Osmor White Spot, etc.
IFT3 3rd IFT, Osmor Blue Spot, etc.
Miscellaneous:
Knob; paxolin; tag strip; chassis about \(6 \times 3 \times 2\) in. deep; \(5 \times 3 \mathrm{in}\). Universal chassis single side; DL32 drive spindle; \(2 \frac{1}{8} i n\). diameter drum; cord; spring; wheels (Home Radio).

Transistors and most other parts are assembled on a paxolin panel \(5 \frac{1}{2} \times 1 \frac{1}{2} \mathrm{in}\). This oscillator/i.f. \(/\) detector strip is later placed in a chassis \(6 \times 3 \times 2 \mathrm{in}\). or \(2 \frac{1}{2} \mathrm{in}\). deep. The chassis supports a drive panel carrying VCI/VC2, so that a cord drive and horizontal tuning scale can be provided.

\section*{Drive Panel}

To avoid metalworking, this is a "Universal Chassis" single side, \(5 \times 3 \mathrm{in}\). This item has a \(\frac{1}{2} \mathrm{in}\). flange all round, and can be bolted to the chassis. \(\mathrm{VCl} / \mathrm{VC} 2\) occupies the position shown in Fig. 3, with its spindle through a clearance hole. Three 4BA bolts secure the capacitor, and these must be short, or have extra washers or other spacers, or screwing them home will damage the capacitor.

Trimmers TCl and TC2 were soldered to tags bolted to the capacitor. A capacior already fitted with trimmers may be used instead. VCl is the larger section, having most plates; VC2 has fewer plates and is to the rear.

Two supports are cut from wood or other insulating material, Fig. 3, and screwed to the drive panel. For easy identification, coloured leads were soldered to the winding tags, as in Fig. 3. The winding is then put on the rod, which is held with string or elastic. Three leads pass through the chassis, to be connected to the oscillator transistor circuits later. Earth returns are completed by the drive panel and chassis.

Two 4BA bolts, with extra nuts, support the scale plate (cut from hardboard \(5 \times 2 \mathrm{in}\).) and small wheels, Fig. 4. Clearance holes are punched in the chassis, for the cord. The spindle drive is situated in


Fig. 2. Wiring and component layout of the receiver.
the middle of the chassis front runner. Arrange the drive, small wheels, and drum so that all are in line. This can be done by moving the drum on the capacitor spindle, adjusting the small wheel spacing nuts, and putting washers or extra nuts on the driving spindle bush, if necessary.

\section*{Drive Assembly}

The cord is taken out through the drum slot, given half a turn round the drum in a clockwise direction, and passed down through the hole. It is given a complete turn round the driving spindle, goes up through the second hole and over the left-hand pulley, across to the right pulley, down to the drum and round this to the drum slot. The ends are tied
with the spring, which goes on the drum projection, under tension.
If necessary, adjust the drum to allow proper 180 degree rotation of VC1/VC2. A small piece of tinplate is clipped on the cord with pliers, and a wire pointer soldered to it. The end goes under and behind the scale plate, Fig. 4, with enough clearance for free horizontal movement. With this diameter drum the actual scale is 3.4 in . long.

\section*{Component Panel}

Figure 2 shows the insulated panel, with all components except R4 one side, and wiring the other side. Coil L2, i.f.t.1, i.f.t. 2 and i.f.t. 3 have can tags which are bent over to hold them in position, and all are joined to the earth or positive line. With

Fig. 3 (right): Method of mounting ferrite rod assembly and tuning capacitor.

Fig. 4 (below): Arrangement of the tuning drive assembly. Note that the drive cord goes through holes in the chassis to the drive spindle-ensure that they have sufficient clearance.


L2, a coloured dot comes
 between pins 1 and 6. I.F. transformers i.f.t.l, i.f.t. 2 and i.f.t. 3 each have pin 4 unused. Other wires must not touch these pins.

The simplest way to avoid errors is to mark each component and lead with coloured pencil, as it is fitted and connected. This shows at once if anything is omitted. The diode and electrolytic capacitor C5 polarity must be as shown. Leads should be quite short and direct, and covered with 1 mm . insulated sleeving where necessary. Connections can be 26 s.w.g. or similar tinned copper wire. Leave a short flexible lead for the negative connection, and another from the diode. Also leave short wire ends projecting from C3, R2 and Trl base.

The finished strip is held in place by two \(\frac{1}{2}\) in. bolts, with extra nuts to leave enough clearance for wiring and R4. One bolt has a tag, connected to the strip positive circuit, so that this is common to the chassis when

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\hline EB91 & 91. & PCC84 & \(21-\) & U282 & \(51-\) & \(20 \mathrm{D1}\) & \(31-\) \\
\hline EF85 & \(31-\) & PCF80 & \(21-\) & U281 & \(51-\) & 20 Ll & \(51-\) \\
\hline EBF80 & 31- & PCF89 & 4/- & U329 & 5/- & 20 P 1 & 51- \\
\hline ECC81 & 31- & PCL82 & 41- & U801 & \(8 / 6\) & 20P3 & \(2 / 6\) \\
\hline ECC82 & \(31-\) & PCL83 & 51- & 688 & \(1 / 8\) & 20 P 4 & \(8 / 6\) \\
\hline ECC83 & \(41-\) & PL36 & \(5 /-\) & 6 BW 7 & 216 & 30PL1 & \(51-\) \\
\hline ECL80 & \(1 / 6\) & PL38 & 6/- & 6K7 & \(1 / 9\) & 30 P 4 & 51- \\
\hline EF50 & 1/- & PL81 & 4/- & 6 K 25 & 51 & 30 P 12 & \(51-\) \\
\hline EF80 & \(1 / 6\) & PY33 & 51- & \(6 \mathrm{U4}\) & \(51-\) & 30 F 5 & \(2 / 6\) \\
\hline EF91 & 9 d . & PY81 & \(1 / 6\) & 6 V 6 & 1/8 & 30 FL 1 & \(51-\) \\
\hline EL36 & 51- & PY82 & \(1 / 6\) & 6 P 28 & 51- & 6/30L2 & 51- \\
\hline EY51 & \(2 / 6\) & PZ30 & 51- & 10 C 2 & 51- & 50CD6 & 61- \\
\hline EY86 & 5/- & U25 & 51- & 10 P 13 & 216 & R19 & \(51-\) \\
\hline
\end{tabular}

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the strip is in position. The brown and black leads are cut and soldered to the correct points, and also the lead from VC2 to C 3 .

\section*{Connections}

A tag strip having one earthed and two insulated tags is bolted to the rear of the chassis inside. The strip negative lead is soldered to one insulated tag, which anchors a black flexible lead passing out of the chassis, and fitted with a plug.

The end of a co-axial lead is prepared, and the outer brading taken to the earthed tag. This forms the battery positive return with the "Pyramid" amplifier. The remaining tag is a junction point for the co-axial cable inner conductor, and lead from diode positive.

Current is obtained from the amplifier by inserting the co-axial and negative plugs. For use alone or with other equipment, the tuner requires a 6 V supply.

\section*{Alignment}

For alignment without a signal generator, tune in any station and adjust the cores of i.f.t.1, i.f.t. 2 and i.f.f. 3 for best volume. A meter in the negative supply lead should also show a drop in current as this is done. Then tune in a transmission with VC1/ VC2 nearly fully open, and adjust TC1 and TC2 for best results. Afterwards, find a station with VC1/ VC2 nearly fully closed, and adjust L2 core, and the position of L1 on the ferrite rod, for best reception. Repeat all the adjustments mentioned, using transmissions of low signal strength, if possible.

If band coverage is unsuitable, this is due to the settings of TC1, TC2, L1 on the rod, and core L2. Should it be found that transmissions around 200 metres cannot be reached with VC1/VC2 fully open, unscrew TC1 and TC2, and re-trim. At the low frequency or high wavelength end of the band (VC1/ VC2 fully closed) coverage can be modified by adjusting the core of L2 slightly, and also moving the winding L1, as required to maintain best reception. The scale can now be calibrated in frequencies or wavelengths by means of a signal generator, or by tuning in transmissions whose frequency is known.

The optimum value for R4 depends somewhat on individual transistors, and is best as low as possible, provided oscillation is not audible when tuning. With some transistors R4 can be omitted. If the tuner is not used with the amplifier mentioned, remember a resistor of about \(5 \cdot 6 \mathrm{k} \Omega\) must be provided in parallel with C 11 , or at the equivalent position in the amplifier.

\section*{TO BE CONTINUED}

\section*{DIARY DATE}

2-5 OCTOBER 1968

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Your
QUESTIONS ANSWERED

\section*{Record Player Interference}

I have a Garrard 1000 record player but find that the reproduction is spoilt by interference from the pickup lead. This interference consists of the distorted reproduction of the record and is still present when the cartridge is disconnected and the turntable unit only is switched on. I have ensured that there are no loose connections.-B. Bishop (Falmouth).

It seems obvious that your amplifier is picking up radiated interference from the motor circuit of the Garrard 1000. First, we would want to know whether the trouble derives from the motor circuit or from the "open" circuit of the pickup. Shortcircuit the pickup input at the amplifier. If this cures the fault, look for common earth connections at the head shell, connector bracket or tagstrip. There should be no common return for the signal via the deck chassis. The deck should be separately earthed and the motor circuit adequately suppressed. If no switching suppressors are fitted try a \(0.01 \mu \mathrm{~F}\) in series with a \(100 \Omega\). The capacitor should be at least 500 V working and the resistor 1 W . A suppressor per switch pole is the normal requirement. The separate earth should not return, in this case, via the amplifier. All signal return leads should return to the amplifier, and not to the deck.
Always watch for hum loops caused by signal earth lines.

\section*{Bandspread?}

I am puzzled by the term bandspread which I have seen used. This appears to apply both to transmitters and receivers. Can you explain please? -A. Davies (Wales).

The term bandspread refers to a system of tuning used in receivers (and transmitters) whereby a small variable capacitor is connected in parallel with the larger main tuning capacitor. This means that for any setting of the main tuning capacitor the small extra capacitor can be used to tune a small band in the region of the main setting. In effect, the extra capacitor permits small increments and decrements in the main tuning.

Another method is to connect a variable voltage capacitor to the tuned circuit controlling the tuning of the transmitter or receiver. This is a capacitor whose extra value depends on the voltage applied to it. Thus by varying a d.c. voltage applied to the capacitor, it is possible to vary its capacitance and therefore the tuning of the circuit. In this instance, a d.c. bias voltage (normally quite low) can be made to alter frequency.

\section*{Radio Club}

I wish to join a radio club, how do I find out where the nearest one is?-F. Mallory (Derby).

Your best bet is to join the Radio Society of Great Britain (R.S.G.B.). Their address is 28 Little Russell Street, London, W.C.1. They will give you the address of the secretary of your nearest radio club.

THE ground plane aerial provides low angle radiation and omnidirectional coverage. It was decided to try such an aerial, as a change from the dipoles, long wires and doublets previously favoured, and after comparing the signal strength of a number of stations who themselves were using ground planes.

For those not familiar with the ground plane, Fig. 1 gives the essential details. The vertical element (usually self-supporting, though it can be wire) is a \(\frac{1}{4}\)-wave long for the chosen band. Each radial is usually a little longer. At least four radials are recommended, more or less evenly spaced round the pole. They also act as guys. The whole is as high as convenient above surrounding objects, and can be fed with \(50 \Omega\) co-axial cable. The cable inner conductor goes to the bottom of the vertical element. and the outer conductor to all the radials.

\section*{Constructional Work}

It was apparent that the whole could be prepared and put up in two or three hours. Clamps were made from \(1 \frac{1}{2} \mathrm{in}\). wide strips of stout sheet metal, Fig. 2. These were fashioned by taking two round pieces of wood, one the size of the pole and the other equivalent to the vertical element, shaping the strips to suit in a vice, and drilling them for 2 in . long \(5 / 16 \mathrm{in}\). bolts. Stout gauge tubing was used for the vertical, and seemed in no danger of collapsing.

The radials were 14s.w.g. aerial wire, looped through two turns of similar wire round the pole. This was drawn tight, twisted, and all joints were soldered. (A large iron is needed.) The co-axial outer conductor is soldered to the ring (radials). The inner conductor is bolted to the vertical element. Joints, and the exposed end of the co-axial insulation, are painted to keep out moisture. The co-ax was stapled a little way down the pole, to take stress off the end.

It was found that the whole could be easily raised into a vertical position, using the method for lifting a long ladder. The bottom end of the pole is pivoted on a post, tied to a strong peg, or placed against something which will not allow movement outwards. The top end of the pole is then raised above the head with both hands. Walking towards
the pole base and simultaneously moving hand over hand along the pole raises it. The pole was tied temporarily to its post, and the radials loosely attached to surrounding objects (a tree, post, and house). The pole was then raised to its higher position (Fig. 1) and the radials drawn tight at 45 degrees to keep it vertical.

\section*{Dimensions}

The aerial erected was designed for near \(14 \cdot 2 \mathrm{Mc} / \mathrm{s}\). The vertical was 16 ft . 6 in . and each of the four radials was 17 ft . long. The standing wave ratio was better than 1.5:1 throughout the \(14-14.35 \mathrm{Mc} / \mathrm{s}\) band. The pole was 18 ft . long, fixed with its bottom


Fig. 1: Elements of the ground plane aerial
end 5 ft . above the ground. For the 10 m or 15 m bands, the length of the vertical, in feet, can be found from \(234 \mathrm{M} / \mathrm{cs}\). The radials are a trifle longer, equal to \(240 \mathrm{Mc} / \mathrm{s}\).

Fecd impedance is low with the radials at 90 degrees to the vertical element (e.g., horizontal) but rises as the radials are sloped downwards. An angle of about 45 degrees is suitable for \(50 \Omega\) co-ax feed, this cable being any length. If circumstances permit, the ends of the radials can be raised or lowered, and the effect on the SWR noted.

\section*{Results}

When first used with a receiver, the ground plane furnished results of about \(5 / 5\) from Australia, South Africa, Philippines, Guiana, and other prominent distant signals. Nearer ranges, such as USA, were around \(5 / 7\) to \(5 / 9\). Closer stations, and Europeans, were generally well up in strength, and one SP (Poland) station also using a ground plane gave a reading of 20 db over \(S 9\).

When transmitting, the ground plane allowed the P.A. to be easily loaded by adjusting the pi tank capacitors. When used with a transceiver in which the method of tuning and particular i.f. caused bad 80 m breakthrough on 20 m with a long wire, the breakthrough disappeared. This was a great benefit when listening.

The best long distance contact reports were \(5 / 5\) with VK and ZL (Australia, New Zealand). This was with 150 watts input. The relative polarisation of aerials, depending in this case on the polarisation used by the other station, seemed to have no bearing on signal strength. On the basis of the reciprocal relation between receiving and transmitting with a given aerial, the ground plane seemed sometimes better and sometimes worse than a dipole and long wire, as would be expected. In any case this relationship does not


Fig. 2: Details of fitting etc. hold for long distance short wave transmission. Equipment used was free from TVI with horizontal aerials. As the home and other local TV aerials were vertical, it was thought that TVI might commence with the ground plane. But in this particular instance TVI was also absent with the ground plane.

The final opinion was that the ground plane was quite a useful aerial to have, and that its actual construction was not a matter of much difficulty. With radials at 45 degrees, such an aerial as that described requires a minimum diagonal space of about \(25-26 \mathrm{ft}\)., or a square of about \(18 \times 18 \mathrm{ft}\).

\section*{A.F. AMPLIFIER MODULE}
-continued from page 392
on the heat sink (Fig. 2) and connecting a millammeter in its place.

Beginning with a low value for R17, higher values are substituted ranging from 10 to \(39 \Omega\), until a current of between 4 and 5 mA is obtained in \(\operatorname{Tr} 5\) at full voltage, care being taken not to warm up the thermistor when soldering. To facilitate the selection of a resistor for R17, temporary lead wires

\section*{\(\star\) components list}

\section*{Resistors:}
\begin{tabular}{|c|c|c|c|}
\hline R1 & 33 k , & R12 & \(470 \Omega\) \\
\hline R2 & 15 k ת & R13 & 33@ \\
\hline R3 & \(2.7 \mathrm{k} \Omega\) & R14 & \(330 \Omega\) \\
\hline R4 & \(100 \Omega\) & R15 & \(180 \Omega\) \\
\hline R5 & \(8 \cdot 2 \mathrm{k} \Omega\) & R16 & \(470 \Omega\) \\
\hline R6 & \(4 \cdot 7 \mathrm{k} \Omega\) & R17 & Seetext \\
\hline R7 & \(1 \mathrm{k} \Omega\) & R18 & \(4.7 \Omega\) \\
\hline R8 & \(2 \cdot 2 \mathrm{k} \Omega\) & R19 & \(4 \cdot 7 \Omega\) \\
\hline R9 & \(680 \Omega\) & R20 & \(4.7 \Omega\) \\
\hline R10 & \(10 \mathrm{k} \Omega\) & R21 & \(4.7 \Omega\) \\
\hline R10 & \(10 \mathrm{k} \Omega\) & & \\
\hline \multicolumn{4}{|l|}{All \(\frac{1}{2}\) watt \(5 \%\) miniature} \\
\hline
\end{tabular}

Capacitors:
\begin{tabular}{ll}
C 1 & \(10 \mu \mathrm{~F} 25 \mathrm{~V}\) electrolytic \\
C 2 & \(100 \mu \mathrm{~F} 16 \mathrm{~V}\) electrolytic \\
C 3 & \(100 \mu \mathrm{~F} 16 \mathrm{~V}\) electrolytic \\
C 4 & \(4,700 \mathrm{pF}\) ceramic \\
C 5 & \(100 \mu \mathrm{~F} 16 \mathrm{~V}\) electrolytic \\
C 6 & \(10 \mu \mathrm{~F} 25 \mathrm{~V}\) electrolytic \\
C 7 & 1000 pF ceramic \\
C 8 & \(64 \mu \mathrm{~F} 40 \mathrm{~V}\) electrolytic \\
C 9 & \(100 \mu \mathrm{~F} 16 \mathrm{~V}\) electrolytic \\
C 10 & 330 pF ceramic \\
C 11 & \(25 \mu \mathrm{~F} 25 \mathrm{~V}\) electrolytic \\
C 12 & \(200 \mu \mathrm{~F} 16 \mathrm{~V}\) electrolytic
\end{tabular}

Transistors:
\begin{tabular}{llll} 
Tr1 & AC156 & Tr4 & AC128 \\
Tr2 & AC156 & Tr5 & AC128 \\
Tr3 & AC127 & Tr6 & AC176
\end{tabular}

Miscellaneous:
Th1 Varite Thermistor type VA1077; Veroboard \(2 \frac{1}{2} \times 3 \frac{3}{4} \mathrm{in}, 0.15 \mathrm{in}\). pitch; 3 transistor cooling clips; \(18 \mathrm{~s} . \mathrm{w} . \mathrm{g}\). aluminium (see Fig. 4); VRY \(100 \mathrm{k} \Omega\) potentiometer
can be soldered in place to which trial resistors can be attached by soldering, the voltage being reduced to zero to enable a new value of resistor to be substituted for R17 and then increased while measuring the collector current of Ti5, taking care that it does not reach an excessive amount.

The chief risk to guard against is an open circuit at R17, and the soldered joints at R17 must remain completely reliable, otherwise electrical failure will occur and replacement of the output transistors will become necessary. In each instance the joints should be tested mechanically before gradually turning up the supply voltage. The total current taken by the amplifier under quiescent conditions is about 27 mA .

\title{
h IOOK qI Infra-RED
}


ADAY by the seaside, that's what the Signals Research Development Establishment offered the press on Thursday, July 11th. When material becomes unclassified, the Establishment makes the information available to industry, and one or two firms were there showing some of their products resulting from this.

To describe the whole tour would be impossible, so let's look at some of the more interesting items. Infra-red was the main attraction and to illustrate advancements in this field, S.R.D.E. promptly set about proving that night or day, you can still be watched.

Two types of system were shown-active, and passive. The active system requires a source of infra-red light but the passive type will "see" a scene in almost total darkness. A moonless night gives adequate illumination for the image intensifier devices.

A pair of binoculars had been converted for use as a communications system, as the photograph and block diagram shows. The image is selected by drilling small holes opposite the prisms. One lens system acts as a transmitter and the other as a receiver, thus one can not only see the other "station" but can talk to them once the binoculars are lined up. A range of around 1 kilometre should be possible with modern Gallium Arsenide devices.

Another fascinating unit appeared to be two blackened cocoa tins mounted on a tripod. These contained a small lens system and a sensitive infra-red detector. Its special trick was that it could detect a man walking in front of it some thirty feet away. "It can tell that the man is not the same temperature as the background," explained our guide. It consumes less than 1 watt too and has a range of 100 metres-no false alarms. The resolution accuracy is 1 degree and the sensitive element which does all the magic is called a thermistor bolometer.

In some situations it might be useful to transmit messages from a fixed transmitter to a receiver situated anywhere in a room, with the provision that the messages must not be intercepted outside. This was demonstrated by illuminating a room with invisible infra-red rays which were modulated with the required signal. Anyone in the room with a suitable receiver could pick the signal up. This might offer useful possibilities in place of the inductive loop principle of radiating a signal around the house. The power used in the demonstration was only 3 milliwatts from a Gallium Arsenide device. Note-the human eye has a maximum tolerance of 60 milliwatts in this area of the spectrum, any increase would be dangerous.

The photograph above shows a pair of infra-red binoculars in use. Below. is a photograph and block schematic of the converted binoculars which form ant optical transceiver. The bottom photograph is of a commercial infra-red viewer for use with an infra-red frght source.



AUND

\section*{No. 1 - AMPLIFIERS, TUNERS, TAPE RECORDERS}

WHAT do we mean by Hi-Fi? The term can be taken literally-faithfulness to the original sound-but if we consider just what we are demanding by this over-simplified approach, we may realise that no equipment, however cleverly designed, would satisfy our requirements. It is necessary to consider the conditions under which we are listening to the reproduced sound. A ninety-piece symphony orchestra in the living room would be a trifle overwhelming!

The first thing we must get clear is the matter of acoustics, the characteristics of the human ear, the range of sounds we can expect to hear (and feel!) their intensity, pitch and effect in combination, together with a consideration of the listening room.

Useful guiding principles have been dealt with by Iain Smith (Five Steps to Hi-Fi. Practical Wireless, May-September) and our present brief is to discuss
some of the equipment parameters that limit the highness of the fi we are able to get.

What we mean by hi-fi is the attainment of a subjective replica of the original sound: the equipment must convey the original sound in its true tonal proportions and not add any coloration of its own. It must be capable of delivering sufficient acoustic power to give us a true impression of the original.

The first thing is to take a look at the sound source from the dynamic point of view. What is the frequency range we wish to reproduce, and what sort of sound power has to be handed? Figure 1 and the frequency scale on the chart give some answers to these questions, but in doing so, raise others, connected with the dynamic range of musical sounds, sound effects (note the "natural" sound scales of Fig. 1) and speech.

Although the ear responds to sounds from below


Fig. 1: Equal loudness curves based on measurements by Robinson and Dadson (reproduced by courtesy of Blandford Press from their forthcoming book \(\mathrm{Hi}-\mathrm{Fi}\) in the Home, by John Crabbe).
\(30 \mathrm{c} / \mathrm{s}\) to above \(15 \mathrm{kc} / \mathrm{s}\) (extended to some \(20 \mathrm{kc} / \mathrm{s}\) in healthy youngsters and limited to about \(10 \mathrm{kc} / \mathrm{s}\) when we get a little grey), our ability to interpret the sounds we hear is modified by the loudness of these sounds.

The vast range in loudness and in frequency demands the utmost care in design if our hi-fi equipment is to handle it successfully. For example, a normal orchestra during a loud passage may produce some 70 watts (acoustic) whereas the solo violin in its quiet passage produces perhaps 0.0000038 watts. This is an intensity ratio of 18 million to one. Taking the square root of this, we find we still have a soundpressure ratio of \(4250: 1\).

We hear logarithmically. That is, our ears respond to proportional changes in sound level, not absolute levels. Every time the sound intensity doubles we hear an equal change in loudness; this applies whether we are listening to a whisper and then one twice as loud or comparing the whumps of a doublebarrelled gun. (In a similar way. our sense of pitch follows a logarithmic law, each frequency doubling representing an octave change, very approximately. See chart.)

\section*{Decibels}

Because of this trick of our hearing, it is more convenient, when talking of sound levels, to express ourselves in decibels. These are ratios, not concrete amounts. It is important to grasp this concept before going any further. It is quite wrong to say that a sound of 60 decibels (dB) was heard. The statement


Fig. 2: Comparison of relative intensities of familiar sounds, showing the audibility limit contours. Compare with Fig. 1.
only acquires meaning when we relate the sound to some known standard. Thus, in Fig. 1 the decibel range is related to a definite sound pressure, i.e. \(0 \mathrm{~dB}=0.002\) dynes per centimetre squared. Then, all increases in sound intensity can be compared with this level, and decreases can also be expressed, as minus quantities. This reference level relates to the threshold of human hearing, which, as the contours of Figs. 1 and 2 show us, change with frequency, being most sensitive at about \(3,000 \mathrm{c} / \mathrm{s}\).

Because of this convenient doubling property of decibels, we can now make the 18 million to one ratio easier to handle, this ratio being 72 dB . Although a front-stall concert-goer may expect even a hundred decibel change when listening to, say, Prokofiev's Fifth Symphony, the average dynamic range our hi-fi equipment is expected to handle is around 60 dB . This 72 dB is the ratio of the softest to the loudest sound, and as we see from Fig. 1, may be between

30 and 102 dB , the 30 dB above threshold referring to the ambient and irreducible noise level in the concert hall.

The unit of loudness, which is the same as the decibel above zero at a frequency of \(1,000 \mathrm{c} / \mathrm{s}\), is known as the Phon, and this can be used as a definite unit-but, just to be awkward, the exact relationship between phons and decibels only applies at this frequency, which is often taken as a reference level.
The decibel has a more useful function than simply expressing ratios of sound pressure. It can be used to compare powers, voltages or current. The difference in level between two powers ( P 1 and P 2 ) is given by \(\mathrm{NdB}=10 \log _{10}(\mathrm{P} 2 / \mathrm{P} 1)\). Voltage and current ratios are expressed as \(\mathrm{NdB}=20 \log _{10}\) (V1/V2) or \(20 \log _{10}(11 / 12)\), because \(P=I^{2} R\) or \(V^{2} / R\), and logarithmically, when we square, we multiply the logarithm by 2 . The 10 and the 20 simply indicate that the decibel, used for convenience, is actually a tenth of the unit, the Bel.
No advanced mathematics are needed to remember the decibel relationship. For practical purposes, it is enough to remember a few key ratios, as, for example:
Voltage: \(2: 1=6 \mathrm{~dB} .10: 1=20 \mathrm{~dB}\).
\(100: 1=40 \mathrm{~dB} .1,000: 1=60 \mathrm{~dB}\).
For combinations, we simply add decibel amounts, thus,
\[
\begin{aligned}
20: 1 & =20+6=26 \mathrm{~dB} \text { (i.e., } 10 \times 2: 1) \\
200: 1 & =40+6=46 \mathrm{~dB}(100 \times 2: 1)
\end{aligned}
\]

Having considered the convenience of decibels, we can begin to use them directly in talking about the dynamics of hi-fi. One of these points is directly related to the loudness phenomenon we have already touched upon. At middle and high frequencies we can judge level differences over a wide dynamic range fairly comfortably, but at lower frequencies, especially below about \(100 \mathrm{c} / \mathrm{s}\), our ears are not so sensitive.
At the lower frequencies, the curves are closer together than at mid and high frequencies and quite large loudness changes produce small stimuli. Which is one reason why the bass end of the audible spectrum is lost first when we turn a gain control down. To overcome this, compensated gain controls are sometimes found, which alter the levels of bass and treble ends of the spectrum in some relationship to the characteristic of the human ear. These loudness controls should be approached with care, as should any form of filtering that attempts to reduce system noise electronically, where, in so doing, some of the basic information may be lost. Nevertheless, listening is a compromise between the ideal and what we can afford, and rumble filters, top-cut controls and loudness compensators may be a necessary evil if we cannot afford to engineer them out!

\section*{Stereo}

For various reasons that have to do with the ambience in the concert hall or studio and the relative deadness (or spirited liveliness) of the domestic surroundings, minus the audience and the original
reverberation, much better effect is gained if we listen binaurally. Stereo systems are to be desired, not only on the grounds of realism, but also for unwanted interference reduction and the proper assessment of complicated waveforms.

Complex waveforms should be studied briefly before we can regard the amplifier practically. Our brains identify sounds by the waveform structure and also by a complicated time-conscious business that can be best described as "hearing the attack".

As an example, let us take the readily identifiable thwack on a bass drum, a sound common both to "serious" and "pop" music. The aural effect that enables us to identify it is the time integral of the pressure levels, a function of the product of level and duration, and is contained in the first fraction of a second. After this, we have a reverberant sound that continues to reassure us, as it were, that it was really a bass drum we heard, not a clap of thunder.

The clash of cymbals is another example, the plucked harp and guitar, the percussive sound of the piano are others. These sounds require equipment responsive to the sudden increase in sound level and the decay that follows it. If you tape record a piano arpeggio and then replay it an immediate sense of loss occurs, because the notes come to us without the attack, or as G. A. Briggs describes it, "like a home-made harmonium suffering from anaemia and groaning in agony". He is not far wrong!

All this means that our equipment must have not only adequate power levels and wide and faithful frequency response, but also a capability of reproducing those sudden attacking sounds--the transients. Of the chain of equipment from source to speaker, the amplifier is the link which can be engineered to the closest tolerances. Having got the heart of the matter right, we can then consider the other parts of the chain-what specifications we require our equipment to have, and why.

\section*{Amplifier specifications}
(1) Frequency response. Accuracy of reproduction depends on the range of frequencies which the system can handle, without distortion. We can detect sound level changes of about 1 dB under good conditionsand this needs very careful engineering.

Even a \(\pm 2 \mathrm{~dB}\) specification over the normal frequency range from \(30 \mathrm{c} / \mathrm{s}\) to \(20 \mathrm{kc} / \mathrm{s}\) is asking rather a lot. It is possible to keep the variation to within

less than \(\pm 1 \mathrm{~dB}\) over a range 40 to \(15 \mathrm{kc} / \mathrm{s}\), at all levels of output power, but to improve upon this specification costs progressively more as the limits extend and the response gets more "flat".

Although we cannot hear the frequencies above, say \(15 \mathrm{kc} / \mathrm{s}\), there are one or two design considerations that apply. First, for an amplifier with a fair amount of negative feedback (see Fig. 3), the bandwidth should extend at least an octave beyond the audio range to maintain stability. Second, to avoid "ringing" caused by a sharp cut-off at the upper end, the curve must slope away at not more than 6 dB per octave attenuation, and this means an extended overall range. And, thirdly, to reproduce square-wave effectively needs a pass-band of ten times the frequency.
(2) Distortion. Even though the response of an amplifier may be reasonably flat, its handling of the signal can still produce a vague feeling of discomfort. Nothing very tangible, until a comparison is made with a better piece of equipment. This effect is often due to a distortion figure higher than the recommended maximum of \(1 \%\).

Non-linear distortion happens when spurious harmonics are added to the original sound because of faults in the reproducing chain. In the amplifier, this may be caused by biasing inconsistencies, especially in transistorised push-pull output circuits, where these are supplied from a badly regulated power source. In fact, this is so important that the latest DIN recommendations are for distortion factors not to exceed \(1 \%\) at full output from \(40-4,000 \mathrm{c} / \mathrm{s}\) for preamplifiers and from full power down to -20 dB over a power bandwidth of \(40 \mathrm{c} / \mathrm{s}\) to \(12.5 \mathrm{kc} / \mathrm{s}\). This latter recommendation is to test transistorised amplifiers at low signal levels, where distortion can arise.

In practice, distortion figures of well below the \(1 \%\) level are attainable, and the magic figure of "Point One" has for years been a valuable (and justifiable) advertising slogan of a well-known British manufacturer. One per cent overall distortion is just detectable by a discerning listener, and the trained ear can


Fig. 3: Block diagram of typical preamplifier showing the extensive use of feedback in fitters and to achieve stable amplification. Gram and tape filtering and equalisation is often in the source equipment but may be incorporated in the preamplifler.
detect half this amount. \(2 \%\) is plainly audible to the music-lover, and \(4 \%\) becomes intolerable to all but the cloth-eared few. By comparison, the cheap transistor radio may produce as much as \(50 \%\) distortion and this will be tolerated!

There are several quite different sorts of distortion. Second harmonic distortion has been virtually eliminated by the use of push-pull output circuits. Figure 4 shows a basic push-pull circuit, each valve (or halfsection of a single valve) handling an opposite phase of the signal.

When one grid is going positive and the anode current is rising, the other is going negative, with anode current falling. Even-order harmonics tend to cancel out. With correctly applied negative feedback, third-order harmonics will also be greatly reduced, and the push-pull stage, with its greater power handling capacity, is extensively employed as a result.

A refinement for hi-fi applications is the distri-buted-load technique. A tapped output transformer from which the screen grids of a tetrode pair are fed gives a characteristic between the triode and tetrode, getting the best of both worlds. This "ultralinear" design requires less negative feedback, reduces total d.c. variations at high output levels and has a relatively smaller capacitative shunting effect at high frequencies as well as less phase shift. Against this, a special transformer is needed, and the position of the tapping point is critical. See Fig. 4 (b).

Most of the non-linearity and phase-shift leading to distortion originates in the output transformer. Experiments to eliminate this led to several interesting circuits, but it was the advent of transistors that made a high power, high fidelity, push-pull output circuit without transformers a tenable proposition.

Figure 5 (a) shows one channel of a stereo amplifier, where we see a complementary pair of transistors in the driver stage perform the function of phase-
changing, eliminating the input transformer, and where the sharing of the signal by a directly coupled Class B push-pull pair of output transistors drives the loudspeaker directly. Figure 5 (b) shows the signal path through this network, with the phase of the signal at any given moment drawn in as a sinewave.
Intermodulation distortion has a very disturbing effect. It arises from sum and difference frequencies of the original tones being produced because of nonlinearity of amplifier response. When a complicated piece of music is being played, with a large number of required fundamentals and harmonics, the extra beat notes of intermodulation distortion can be quite intolerable. DIN specifications require that intermodulation distortion be assessed separately and must be less than \(3 \%\) maximum when two test signals of \(250 \mathrm{c} / \mathrm{s}\) and \(8 \mathrm{kc} / \mathrm{s}\) are simultaneously applied at an amplitude ratio of 4:1.

From the practical standpoint, there are two identifying features of IMD (Inter-Modulation Distortion) which can be spotted when, for example, a choir accompanied by an organ, has been recorded. The long-term IMD produces a blurring effect as the harmonies become more complex, and the beat-note distortion due to mingling of two fundamentals close together gives a low frequency "blasting" effect, most noticeable with organ music.
The foregoing spec. is quite generous, and most manufacturers of high-fidelity equipment would aim at a figure of \(1 \%\) IMD.
(3) Transient response. Transients are short-term peaks of sound. We have already seen that the attack is important in identifying a sound and aiding stero location. Unless an amplifier has good transient response, the initial \(10-50 \mathrm{~dB}\) increase in a period as short as \(100 \mu\) S will be flattened out, or, worse, will give rise to a "ringing" of circuits and spurious distortion effects.

Fig. 5 (right): Transistor Class \(B\) push-pull transformerless output stage (a); signal paths of push-pull amplifier showing phase reversal effect of complementary push-pull driver pair.

Fig. 4 (below): Basic pushpull circuit (a), basic ultra linear circuit (b).

(b)

Figure 6 shows the simplified response curves taken from four amplifiers with similar overall specifications but different transient response characteristics. A square wave applied to each amplifier produces an output as shown in A, B, C, and D, and the curves resulting from this distorting effect are shown on the graph.

Amplifier A has a slight overshoot due to a small peak at about \(60 \mathrm{kc} / \mathrm{s}\) (well above the audio range), and \(B\) has a larger peak.

Amplifier A might be acceptable, but would possibly be triggered off by switching transients into some instability, especially if loudspeaker loading was poor. Amplifier B would certainly sound harsh, and transients would cause instability. The frequency


10kc/s square wave input
Fig. 6: Four different frequency response curves and the effect on a \(10 \mathrm{kc} / \mathrm{s}\) based square wave of each of the four amplifiers. (Transient response is affected by the overall response curve of the amplifier).
response of C tails off too rapidly at the higher end, although still only about 3 dB down at \(10 \mathrm{kc} / \mathrm{s}\), which some mid-fi makers consider acceptable. An applied square wave would come out something like the peculiar shape below, and the sound from it would be dull and lacking in attack. In a piano arpeggio, the notes would tend to run together instead of being individual and distinct as they should be with D.

Transient response is affected by some tone (and even gain) control circuits, and for this reason testing is done at different levels and frequencies. The effect on the response curve of different tone control circuits can be seen in Fig. 7. The passive type of tone control has the effect of "hingeing" the response about a \(1 \mathrm{kc} / \mathrm{s}\) centre, but with the Baxandall type the boost and cut is initially confined to the ends of the scale.
This is most useful when the lower bass frequencies need boosting without affecting the \(300-500 \mathrm{c} / \mathrm{s}\) region. Where slope filters are fitted--providing topcut fairly steeply to reduce distortion from the source, an ordinary passive tone control may be enough, but a combination of both is better. See Fig. 8.
(4) Signal-to-noise ratio. Noise can consist of hum, rumble (from turntables, etc.), hiss (from tape or preamplifiers) and random crackles, etc. While the power amplifier can be engineered to a very good specification, noise from preamplifiers, where the switching, tone controls and filters will be situated, is still difficult to eliminate.

(a)


DIN specifications for noise level stipulate better than -50 dB for preamplifiers at the nominal input signal amplitude and similarly for power amplifiers up to 20 watt rating, with a 50 dB figure also for a 100 mW output, which is a test for transistorised amplifiers.

The minus 50 dB indicates that the level is measured in decibels below the maximum output of the equipment, this maximum determined by the specified distortion figure. A 50 dB signal-to-noise ratio (which is the same thing stated positively) is generous. Most amplifiers would be better than 60 dB , and a good hi-fi amplifier would have a \(\mathrm{S} / \mathrm{N}\) ratio of 70 to 100 dB , much better than the figures of the source material, whether from tape, disc or radio.

As an example of the expected figures, we quote the requirements of one notable reviewer for hum and noise content of a hi-fi amplifier: these are minimum figures from various sources, each terminated by the appropriate load resistor.
\begin{tabular}{|lcc|}
\hline \multicolumn{1}{|c|}{ Source Input } & Sensitivity & \begin{tabular}{c} 
Hum and Noise \\
(rel. 10W)
\end{tabular} \\
Tape Head & \(2 \cdot 5 \mathrm{mV}\) & -48 dB \\
Magnetic pickup & 3 mV & -55 dB \\
Crystal orceramic & 50 mV & -60 dB \\
\begin{tabular}{l} 
pickup \\
Radio
\end{tabular} & 200 mV & -65 dB \\
\hline
\end{tabular}

Noise figures may be weighted or unweighted. This means that in testing some account has been taken of the peculiarities of our hearing apparatus. The ear is less tolerant of some kinds of noise than others, and will put up with a lot of mid-frequency components (where the ear is most sensitive). The main noise spectrum of transistor amplifiers lies in the mid-region and a weighted figure gives a better idea of transistor amplifier performance, subjectively.
(5) Power Output. Here we have what may be a stumbling block to many readers, because of the different methods of measurement-and the habit of some manufacturers to give a bald " \(X\)-watts Output" statement, without saying whether this is r.m.s., continuous sinewave, music power or what-have-you. Be wary of such specifications !


Fig. 7: Typical curves showing the effect of (a) passive and (b) active-in this case Baxandall-tone control circuits.


Fig. 8: Passive (a) and B-axandall (b) tone control circuits as used in practice.

DIN specifications recommend at least 10 watts mono and 6 watts each channel stero, with a capability of producing sinewave signals of \(1 \mathrm{kc} / \mathrm{s}\) for a period of 10 minutes. Power output figures are related to a given level of distortion.

In this country, an r.m.s. figure is generally stated, referring to a maximum continuous output power at the specified distortion figure.

Peak power ratings are more generally quoted by American and Japanese manufacturers. We find the term "music power" in use. The IHFM definition states that: "Music power shall mean the greatest single frequency power that can be obtained without exceeding the total rated harmonic distortion when the amplifier is operated under standard test conditions, except that the measurement shall be taken immediately after the sudden application of a signal and during a time interval so short that supply voltages have not changed from their no-signal values."

To begin with, this argues a well-regulated power supply, or an external stabilised power supply. The peak power is obtained by doubling the power rating. Figure 9 shows a half-cycle of a sinewave, with the comparison between peak and r.m.s. values. The effective continuous voltage is 0.707 times peak voltage and continuous power is half \(\left(0.707^{2}\right)\), the peak power. Music power figures may give a false impression, often being some \(30 \%\) above sinewave ratings. Care must be taken when studying these specifications.

Power bandwidth is also important when considering specifications. This relates to the frequency range lying between the extremes where power output falls by 3 dB -or a half. Figure 10 compares the response related to power output of two quite dissimilar amplifiers. Both have a peak power handling capacity of 10 watts, and were both, in fact, sold as such. But whereas A gives a half-power figure at \(15 \mathrm{c} / \mathrm{s}\) and \(30 \mathrm{kc} / \mathrm{s}, \mathrm{B}\) is restricted to \(60 \mathrm{c} / \mathrm{s}\) at the lower end and only \(10 \mathrm{kc} / \mathrm{s}\) at the upper end. Definitely midfi!

Power output at the upper end is important for good transient response and the half-power point should be \(30 \mathrm{kc} / \mathrm{s}\) or above for a good amplifier. At the lower end, half-power at \(40 \mathrm{c} / \mathrm{s}\) is desirable, and at \(20 \mathrm{c} / \mathrm{s}\) even better, although the fundamentals of few instruments go down so far. The piano and contra-bassoon and the lower strings of the harp go below \(40 \mathrm{c} / \mathrm{s}\), but as the second and third harmonics are greater than the fundamental, losses are not too obvious.

But the organ can only be reproduced in the region where its music is "felt" rather than heard, with an amplifier (and accompanying system) whose half-

power rating is \(20 \mathrm{c} / \mathrm{s}\) or below. This is where transformerless amplifiers have a decided advantage.
(6) Stereo separation. Although some authorities maintain that this specification is not so important, because a separation as poor as 10 dB will still give a good stereo impression if the frequency range is wide enough and the transient response is good, the higher the separation the better.

DIN recommendations require crosstalk between stereo channels to be better than -50 dB at \(1 \mathrm{kc} / \mathrm{s}\) and -30 dB between \(250 \mathrm{c} / \mathrm{s}\) and \(10 \mathrm{kc} / \mathrm{s}\). Breakthrough between inputs should be -50 dB or better at \(1 \mathrm{kc} / \mathrm{s}\) and -40 dB or better between \(250 \mathrm{c} / \mathrm{s}\) and \(10 \mathrm{kc} / \mathrm{s}\). Limiting factors are often sources, such as pickup cartridges and the discs themselves.

Tapes can achieve a better preparation, provided the tape recorder is properly adjusted and the preamplifier correctly designed. V.H.F. tuners should achieve a -30 dB figure. For an amplifier of any pretensions to quality, -60 dB should be aimed at.
(7) Sensitivities. Nominal sensitivities should relate to the specified output. When an input goes through a volume control, non-linear distortion should be less than 1 dB when inputs are 12 dB above nominal levels. Magnetic pickup: loading, \(47 \mathrm{k} \Omega\), input sensitivity 5 mV .

Crystal and ceramic cartridges: sensitivity is less, often 50 mV or so, and impedances much higher. But it is possible to attenuate externally, or to apply a ceramic cartridge as a pressure-gradient source.

Radio tuner: many Continental tuners are designed to match a high sensitivity, \(47 \mathrm{k} \Omega-100 \mathrm{k} \Omega\)


Fig. 9 (left): Peak power and r.m.s. values with output volts related to time and one half of the sinewave test signal shown.

Fig. 10 (right): Half-power figures are useful to indicate the power capability of an amplifier, where a peak figure is practically meaningless. Power bandwidth relates response to output power.
input whereas British tuners generally deliver a higher output into \(500 \mathrm{k} \Omega\) or so. But matching presents few problems. Outputs are specified as 1 volt across \(47 \mathrm{k} \Omega\) for matching or a preamplifier to a power amplifier, and for connection to a tape recorder should be \(0 \cdot 1 \mathrm{mV}\) to 2 mV for every \(1 \mathrm{k} \Omega\) of resistance from \(1 \mathrm{k} \Omega\) to \(50 \mathrm{k} \Omega\). There are many practical variations of this and selection should be made with care.

Power output matching is recommended at 4 and \(16 \Omega\), (our normal 3 and \(15 \Omega\) loudspeakers suit these requirements). It should be remembered that transistor amplifiers tend to deliver greater power into a lower impedance, but are easily damaged by shunting with too small a matching impedance-the opposite to a valved amplifier, which dislikes an open-circuit.

Common practice with good quality equipment is to protect against overloads, short-circuits and power failures, but discussion of this is beyond the scope of these notes.

\section*{RADIO TUNERS}

The radio tuner is nowadays an integral part of any hi-fi set-up. Many good programmes are available, and broadcasting quality on v.h.f. is capable of as fine results, provided reception conditions are sufficiently good. The tuner is, basically, a radio set without a power amplifier (and in some cases, without


IOTIIT suppaneli
a "line-of-sight" pattern, about 50 miles or so radius, a network of stations has been (and is still being) built up by the BBC to give full population coverage.
But it is necessary to employ a good aerial to get the best from v.h.f., not only because of signal strength, but to eliminate multipath effects (ghosts) which cause a buzzing background noise or distortion and tuning in some cheaper equipment.
It is also false economy to make do with an inefficient aerial on the grounds of noise suppression. Electrical interference is impulsive in nature, varying the amplitude of the signal. The f.m. signal is transmitted at a constant amplitude and thus we can eliminate impulsive noise by limiting the amplitude of the signal within the tuner, "chopping the peaks off", so to speak. But to get the best limiting effect, it is necessary to drive the tuner as hard as possible. Overloading is protected by automatic gain control circuits, which are a common feature.

Multi-element aerials give greater gain and better a power supply unit also), but, because it is specially designed for the purpose of matching other hi-fi equipment, the standards to which it may be designed and constructed will be much more rigorous.
A.M. and f.m. tuners, and some with combination of a.m./f.m. facilities, are available, but for the purposes of high fidelity reception we should consider f.m. The only reason for having broadcast bands on our tuner would be the need to reproduce stations not available on the v.h.f. band-and as the quality of programme from such stations is often dubious (from a technical point of view !) serious consideration should be given whether it is cheaper to buy a good v.h.f. tuner and a separate cheap radio for the odd a.m. broadcast, or pay a lot more for a combination a.m./f.m. tuner.
With a.m. the bandwidth is limited by factors beyond the broadcaster's control. Broadcast bands are very crowded; interference is rife. Transmissions can fade due to weather conditions and with time of day. Locally generated noise due to electrical apparatus presents the same pattern to the receiver as the modulating signal, and cannot thus be reduced without some curtailing of vital programme information. Highly selective and highly sensitive receivers can be designed, and are on the market at high prices, but these are more attractive to DX listeners than to the hi-fi enthusiast whose prime aim is high quality of sound. Restricted frequency response and high noise level are the two main drawbacks to a.m.

\section*{V.H.F. Band}

There is more room in the v.h.f. band. The frequency range transmitted can be higher and, in fact, \(15 \mathrm{kc} / \mathrm{s}\) is reckoned to be available on many broadcasts, although land-lines between studio and transmitter and between stations limit the upper frequencies somewhat. Because the service range of the higher frequency transmissions extends mainly over


Fig. 11: Block diagram of f.m. tuner with decoder expanded to show operation (see text).
directivity and provide the correct match to the aerial input terminals of the tuner, which may be 300 or \(75 \Omega\), the latter favoured by most British manufacturers. Even a simple H aerial, erected out-of-doors and at an adequate height, is sufficient for most purposes. But although an indoor aerial may appear to give enough signal to drive the tuner, spurious signals can occur as it is "shadowed" by people moving about, or by reflections from internal plumbing, etc.

A loft aerial is a partial solution but housewiring, water-tanks and even the difference in field-strength of the signal shaded by a wet or a dry roof can give rise to reception changes. Height of aerial is a vital factor. Doubling the receiving aerial height is equivalent to multiplying the transmitter power four times !

Preamplifiers can be used to boost the f.m. signal, but there are drawbacks. The preamplifier must be of equal bandwith to the aerial, it must not of itself contribute noise to the system, and it should, for best effect, be mounted as near the aerial as possible, i.e. at the masthead-which raises powering problems.

It cannot contribute to directivity and, if the aerial is picking up noise, will amplify both noise and sig-
nal together. The solution, in all cases, is to capture as good a signal as you can with as good an aerial as you can erect, taking care over its siting.

Matching to the tuner is no difficulty. Normal coaxial cable is self-screening and has a nominal \(75 \Omega\) impedance. Where a tuner only has a \(300 \Omega\) input, a match can be made by connecting the outer braid of coaxial to the tuner chassis and the inner to one of the terminals, leaving the other unconnected. This gives an impedance match of \(1: 4\) and uses the tuner input circuit as a 1:2 transformer.

Three hundred ohm cable (actually 240-300 \(\Omega\) rib-bon-type) is not screened and its installation presents difficulties. It must be kept parallel at all times, should not run alongside guttering, pipes, etc., and should be spaced evenly away from all securing points. Aerial polarisation is horizontal for f.m. transmissions.
a \(19 \mathrm{kc} / \mathrm{s}\) pilot tone) and amplified, doubled and fed to the synchronous detector is also applied to the matrix. This translates back into the original A and B signals of the stereo broadcast for application to the main amplifier.

The important thing to note is that both the \(19 \mathrm{kc} / \mathrm{s}\) pilot tone and the \(38 \mathrm{kc} / \mathrm{s}\) carrier are available at the output from the tuner. This is no problem as far as radio reception via the hi-fi amplifier is required, but becomes a difficulty when tape recorders are part of the link-up. Either or both of these can beat with the tape recorder oscillators, giving rise to spurious notes within the audio spectrum.

Suppressors need to be fitted, usually in the form of notch filters, to eradicate these continuous tones. Many tape recorders in the high quality bracket already have these. Regrettably, not all are as effective as they might be, and as stereo broadcasting

\section*{Stereo}

Stereo broadcasts thoroughly justify the use of f.m. tuners with hi-fi gear, although these are at present limited to part - time broadcasts from only six stations. The BBC is extremely keen on high quality broadcasting, and the stereo network will grow. But although the same frequencies are used, the extra information needed

for stereo reception is conveyed by a suppressed sub-carrier. This, with the pilot tone necessary to operate the decoder, means that a stereo signal has to be some 20 dB stronger than a mono signal to obtain a comparable signal-to-noise ratio. Another point in favour of a good aerial installation.

The important factor in drawing up the specifications for stereo broadcasting was the need for compatibility. As with colour TV transmissions, the new service had to coincide with, rather than supplant the old. Receivers of mono transmissions must be able to pick up the stereo broadcasts in mono and stereo receivers had to accept the mono transmissions still being broadcast. The stereo signals contain all the mono information plus stereo information for the individual channels and a pilot tone to enable the decoder to work.

\section*{Typical Tuner}

Figure 11 shows the block diagram of a stereo tuner, with the decoder section expanded to illustrate its operation. Up to the detector output, it is a normal mono f.m. tuner. From the detector the M and S signals plus the \(19 \mathrm{kc} / \mathrm{s}\) pilot tone are passed to the decoder. The \(M\) signal is the sum of the individual channel signals at the transmitter, and the S signal is the difference between them, with the \(38 \mathrm{kc} / \mathrm{s}\) suppressed so that only the sideband signals are transmitted. All three signals, M, S and pilot tone, modulate the carrier simultaneously.

In the decoder, Filter 1 selects the mono signal and passes it to the matrix. The \(S\) signal is passed via. Filter 2 and is passed to the synchronous detector where the \(38 \mathrm{kc} / \mathrm{s}\) carrier, picked up via Filter 3 (as
increases, we shall find ourselves carrying out suppression experiments on all classes of equipment. Specifications for tuners and tuner-amplifiers take this into account.
It should be remembered that not all f.m. tuners can be converted to stereo simply by the addition of a decoder. Although most foreign-made tuners already have decoders, because they were originally designed for use in countries where f.m. stereo broadcasting has been an accepted thing for many years, in the UK we find that manufacturers have geared production to demand and are only lately speeding up decoder production.
Cost of an additional decoder, including fitting, may put twenty pounds or so on the price of a tuner, and this factor should be considered when choosing. Again, some tuners will only accept the decoder designed by the same makers-unless one happens to be a handy constructor.
DIN recommendations relating to tuner-amplifiers are as follows: The standard is based on an aerial input of 1 mV across \(240 \Omega\) and an a.f. output 6 dB below full volume except for distortion factor measurements.
Frequency range must be at least \(40 \mathrm{c} / \mathrm{s}\) to \(12.5 \mathrm{kc} / \mathrm{s}\) with permissible deviations (relative to \(1 \mathrm{kc} / \mathrm{s}\) ) of \(\pm 4 \cdot 5 \mathrm{~dB}\) from 40 to \(50 \mathrm{c} / \mathrm{s}, \pm 3 \mathrm{~dB}\) from \(50 \mathrm{c} / \mathrm{s}\) to \(6 \cdot 3 \mathrm{kc} / \mathrm{s}\) and \(\pm 4 \cdot 5 \mathrm{~dB}\) from 6.3 to \(12 \cdot 5 \mathrm{kc} / \mathrm{s}\).

Channel balance must be not worse than 6 dB from 250 to \(6 \cdot 3 \mathrm{kc} / \mathrm{s}\) and not worse than 9 dB when balance control is fitted giving an adjustment of at least 8 dB .

PLEASE TURN TO PAGE 418

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-continued from page
416
Distortion should be less than \(2.5 \%\) at \(1 \mathrm{kc} / \mathrm{s}\) with \(40 \mathrm{kc} / \mathrm{s}\) deviation--when the same signal is applied to each channel-with a power bandwidth of \(40 \mathrm{c} / \mathrm{s}\) to \(12 \cdot 5 \mathrm{kc} / \mathrm{s}\), the output correctly terminated.

Crosstalk: permissible figure is given as 24 dB at \(1 \mathrm{kc} / \mathrm{s}, 18 \mathrm{~dB}\) from \(250 \mathrm{c} / \mathrm{s}\) to \(6.3 \mathrm{kc} / \mathrm{s}\) and 14 dB from 6.3 to \(10 \mathrm{kc} / \mathrm{s}\).

Signal-to-noise ratio (unweighted) relative to 100 mW (mono) and \(2 \times 50 \mathrm{~mW}\) (stereo) from systems up to 20 W output should be better than 40 dB from \(40 \mathrm{c} / \mathrm{s}\) to \(15 \mathrm{kc} / \mathrm{s}\) with input of \(1 \mathrm{kc} / \mathrm{s}\) at \(40 \mathrm{kc} / \mathrm{s}\) deviation initially. Overall noise is given as better than 50 dB for both mono and stereo systems between \(40 \mathrm{c} / \mathrm{s}\) and \(15 \mathrm{kc} / \mathrm{s}\).

Pilot tone \(\mathbf{S} / \mathbf{N}\) ratio, when measured selectively at \(19 \mathrm{kc} / \mathrm{s}\) and \(38 \mathrm{kc} / \mathrm{s}\) should be equal to, or better than 19 dB and 29 dB respectively. This is with the input signal of \(1 \mathrm{kc} / \mathrm{s}\) deviated by \(67.5 \mathrm{kc} / \mathrm{s}\) and a level of 1 mV into \(240 \Omega\).


Fig. 13: (a) short-term flutter on a longer term wow may not be easily measureable by simply reading peak values; (b) pen recording showing the recurrent "pips" in waveform coinciding with mechanical vibrations in the tape transport.

TAPE RECORDERS.
Here we introduce another factor into the hi-fi specifications-mechanical variations. Most readers will be familiar with the terms "Wow" and "Flutter", but a few words on the exact meaning of the terms may help to explain the specifications.
Speed variations take three forms: long-term stability, slow variations of the one to ten cycles per second periodicity and rapid variations that may extend well into the audio range. The effect of longterm speed variation may not be very noticeable, except on exchange of tapes or other methods of direct comparison, as pitch is the only thing affected to any extent, unless the variation is so severe as to be immediately audible.

Slow variations in the form of wow can be very obvious indeed, especially on piano and flute notes, where the speed variation causes pitch wobble of a quite distinctive nature. Our hearing mechanism is sensitive to this kind of change. As little as \(0.15 \%\) is quite evident on piano notes around the midfrequency region. In fact, the ear is a very good judge of wow, and instruments capable of making a comparable assessment are quite expensive pieces of laboratory equipment. The DIN recommendation for a combined wow and flutter content of less than \(0.2 \%\) peak-to-peak would be too generous for most hi-fi enthusiasts.

But much depends on the basic speed of the tape transport. Wow is much worse at slower speeds. High quality recording is generally done nowadays at \(7 \frac{1}{2} \mathrm{in} . / \mathrm{sec}\)., even though much pre-recorded material is available at \(3 \frac{3}{4} \mathrm{in} . / \mathrm{sec}\). A wow and flutter figure of \(0 \cdot 15 \%\) should be regarded as the maximum at \(3 \frac{3}{4} \mathrm{in}\)./ sec. for any sort of quality to be gained.

Flutter is the short-term variation, which can be caused by a number of small discrepancies in the tape transport system and pressure pad arrangement, and shows itself as a harshness much like the sound of intermodulation distortion in the amplifier. The two effects, though measurable separately are more often lumped together as overall speed variations, and, indeed, more than one authority is now doing this and using the blanket term "wobble".

Wobble may be measured as an r.m.s. value-sinewave signals being used for the test-or as a peak-to-peak value, which may be easier to understand, but can be misleading unless stated as such. Peak-to-peak values will be twice the peak value, and r.m.s. values are 0.707 or roughly two-thirds of peak value. It is therefore necessary to know, before buying, whether the wow and flutter figure was made under peak, peak-to-peak or r.m.s. conditions, and at what speed.


Fig. 14: Frequency response curve of amplified tape recorders are specified to fall within the tolerance limits shown

In the DIN specifications, variations between medium and average speeds over 30 seconds are not to exceed \(\pm 1 \%\). Reproduction range of tape equipment should cover the 40 to \(12.5 \mathrm{kc} / \mathrm{s}\) range, with the response lying within the shaded area of Fig. 14. Full amplitude, measured at \(333 \mathrm{c} / \mathrm{s}\) is reached by a cubed distortion factor of \(5 \%\), at peak recording level, and the signal-to-noise ratio, referred to peak recording level should be better than 45 dB .

For stereo machines, track separation should be 60 dB and channel separation 25 dB . Erasure should be 60 dB below peak recording level.

Amplified tape recorders, i.e. decks with preamps, are given additional specifications. These may be summed up as the foregoing, plus:

Stereo crosstalk to be at least 24 dB at \(1 \mathrm{kc} / \mathrm{s}\) and at least 21 dB between 250 and \(10,000 \mathrm{c} / \mathrm{s}\). Signal/noise ratio better than 41 dB (which is remarkably modest).

Output power is as for amplifiers, 10W mono and \(2 \times 6 \mathrm{~W}\) stereo, the power to support a sinewave signal for at least 10 minutes ( \(1 \mathrm{kc} / \mathrm{s}\) ).

Normál input and output sensitivities and impedances on tape recorders are much as amplifiers, except that a "line" output of 775 mV at \(100 \mathrm{k} \Omega\) is becoming widely used and many variations of diode input between manufacturers-to match units of their own make-have rendered any standardisation impossible.


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0VER the years a large amount of material has been published in this magazine concerning optical communication systems. All of these systems have been concerned only with short range work, i.e. up to 30 yards or so using small lenses, and the writer decided to see what could be done about transmitting over longer distances. Two types of transmitter were tried: neon bulb and torch bulb. A third type will be tried some time in the near future using infra-red, obtained from a Gallium Arsenide electroluminescent diode, type CAY12. At present this is available from Mullard Ltd., but only through the trade.

\section*{Neon Transmitter}

The circuit chosen for this unit was that published in the April 1967 edition of Practical Wireless and designed by H. L. Mason. This comprised a one valve two stage amplifier using an ECC82 valve. The anode load was a neon bulb plus current limiting resistor. The transmitter was built into a small aluminium case \(4 \times 4 \times 3\) in. high which also housed the power unit. A co-axial socket was mounted on the top with the on-off switch and two wander plug sockets were mounted on the side for the neons. WARNING: High, voltage is present at these sockets.

When the unit was being set up a small transistor multivibrator was used to provide a tone, but when speech was required a crystal microphone was used with a pre-amp. If the quality of the speech is of no concern then a carbon microphone plus battery and transformer could be used in which case the pre-amp would not be needed. The neon is capable of providing high quality sound and has a very good high frequency response. The current taken by the neon is about 2.5 mA with no input, falling to about \(750 \mu \mathrm{~A}\) when fully modulated. The

unit can provide power for up to four neons plus resistors in parallel, but it is difficult to use them all due to the small focal point of any reflector.

The reflectors used in the prototype were approx. 15 in . in diameter and parabolic in shape. In theory a parabolic reflector will produce a parallel beam of light if a point source is placed at the focus. This was far from true in the reflectors used, but even so there was little to be gained from having more than one neon.

The neons were mounted in old discarded ball-pen tubes of the 'Biro' type. The ink tubes and end stops were removed, and the constriction at the end removed with a hack-saw. The neons used, made by MK, had a \(270 \mathrm{k} \Omega\) resistor soldered on to one lead, which was replaced by a \(100 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W}\) type. There is not much space available in the tube so this could profitably be replaced by two \(47 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W}\) types in series. During this operation small pieces of sleeving were slipped on to the wires to insulate them. Two leads are then soldered on and the assembly mounted in the tube. If the unit is operated with no input then the resistor will run a little warm but this is nothing to worry about. The neon(s) were mounted at the focus of the reflector with standard laboratory type clamps and bosses.

\section*{Torch Bulb Transmitter}

This unit used the amplifier from the neon transmitter with the neon replaced by an output transformer. A series resistor was also included to reduce the no signal current to below the limit for the valve. It was not considered worthwhile to change the bias on the valve because the author did not, at first, anticipate using the system for more than test purposes, and it worked well like this anyway. The output transformer was a small low power pentode type which was to hand. Most types should work in this circuit. Here again watch out for high voltages on the primary tags.

The bulb unit was the main beam part of a battery lantern. The connection between the centre pip on the bulb and the contact spring on the torch was broken by inserting a piece of p.v.c. tape, and two wires were soldered, one to each of the connections.

Fig. 1: Three of the receiver circuits tried by the author; (left) photo-voltaic cell, (centre) emitter follower transistor circuit and (right) phototransistor connected to a simple amplifier.

These wires were connected at the other end to the secondary of the transformer. As the lantern was powered by a six volt battery, the input did not fully modulate it. Even so the modulation could be seen, and the extra brightness helped to carry the signal.

The torch system is totally unsuitable for carrying music due to the very bad frequency response of the bulb (caused by the fact that it takes a finite time to heat up and cool down each cycle), but this does not matter where speech is concerned as intelligibility is the main requirement. The lantern gave a good beam which was only about 15 degrees across.

\section*{Reflectors}

The most important link in the chain is the pair of reflectors or lenses used. Reflectors are to be preferred to lenses of the same diameter at the transmitter, as a larger solid angle is covered and thus a larger proportion of the light is collected and aimed at the receiver. There is no reason why a lens should not be used at the receiver, as it is only the area covered that is of importance, and a lens will probably have a better focal point than a reflector. The problem with lenses is that large ones are hard to come by, the best are probably those from old signal lamps. If reflectors are to be used then the best easily obtainable types are WD surplus searchlight reflectors, or those removed from a car headlamp.


Fig. 2: Circuit of the torch bulb transmitter.

\section*{The Receiver}

Several types of receiver were tried: photovoltaic cell connected across the input of an amplifier. This was not very sensitive and was discarded. Emitter follower circuit using a photo-resistive cell as part of the biasing network. This was fairly sensitive but was changed for experiment's sake to a phototransistor itself connected in a simple amplifier circuit. This was not found to be as sensitive as the final circuit chosen which was a battery, resistor, and photoconductive cell in series. When light falls on the cell its resistance decreases causing the current through it, and thus the resistor to increase. Thus the voltage drop across the resistor increases and the input
signal is formed. The cell used was an OC71 transistor with the opaque paint removed, connected by the emitter and collector leads with the base left unconnected. A \(3.3 \mathrm{k} \Omega\) resistor was used with a 4.5 volt battery which passed a few milliamps through the cell, not enough to exceed the rating.

The input signal was fed to a pre-amplifier (Eagle type EM-3) and then to a Heathkit 10 watt amplifier. The battery used in the prototype to power the photo cell was separate from that used for the preamp, but no doubt the same one could be used.
When the system was assembled in the lab. the receiver was completely overloaded, producing distortion. In the case of the neon transmitter acoustic feedback occurred whenever the microphone was connected, even though it was thirty feet away from the loudspeaker. To prevent feedback the gain of the amplifier had to be turned right down. When the torch bulb transmitter was used feedback of the normal sort was avoided. Instead, a low pitched gurgle was produced.


Fig. 3: Final receiver design used by the author.
The system was put into operation across a play-ground-a distance of about 75-80 yards, between a lab. and a workshop. The beam had to pass through two windows en route. When the neon was used it took some time to align the reflectors for optimum performance. This was due to the beam being red, and also spread over a large area of reflector. With the reflectors used it was possible to utilise about \(2 / 3\) rds. of the area at any one time. The only efficient way to increase the power from this transmitter is to use more than one neon, each neon having its own reflector. As it was, plenty of power was available to operate a telephone handset. It was felt that this was the useful limit of the neon as regards distance.

With the bulb connected greater power was available at the receiver and the distance could be increased to 300 feet or more. It was found much easier to line up the receiver due to the beam being concentrated from a point source. To achieve the greatest distances the per cent modulation of the beam should be increased.

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\section*{Radio Eireann}

In the June 1968 issue of Practical Wireless in "Your Questions Answered", a Mr. Browne of London, N.W.1, requested information \(r e\) better reception of Radio Eireann on 530 metres.

As I am a regular listener of this station I feel that I can help Mr. Browne with his problem. Briefly, the effective improvement I use is to resonate a short indoor aerial to the desired frequency by means of a series inserted coil wound on a ferrite rod and placed beside the receiver, effecting coupling to the receive aerial.-D. Walsh (13 Sixth Avenue, Chelmsford, Essex).

\section*{P.O.P.}

Reader's letter in issue No. 738, Postal Order Problem, J. Martin, Halifax. I had a postal order returned to me, so here is the procedure I carried out. I took the P.O. and the counterfoil to the office of issue. Postal order signed in the usual manner in the presence of the postmaster, money refunded. If unable to contact Post Office of issue, contact the Head Postmaster for instructions.-J. Wright (Co. Durham).

\section*{Things that go Bump}

Re Mr. S. Pinder's letter "Your Questions Answered" headed Audio Thump.

I feel that his trouble does not lie in the power supply, as he suggested, but (assuming his amplifier has a transformerless output stage) in the loudspeaker decoupling capacitor.

The remedy in this case is to connect two large capacitors of similar value across the power supply, and to connect the speaker to the centre tap. This simple modification provides an artificial a.c. centre-tap across the supply lines, and has completely cured a similar problem in my own ampli-fiers.-G. Fecitt (Lancs.).
I have read this month's "Questions Answered" and I wish to offer a few words on the Audio Thump from S. Pinder.
This thump, has, I suppose, nothing whatever to do with the
power supply in any way. Almost without exception transistor power amplifiers have this thump which occurs only when switched on after being off for some time. Most amplifiers are designed for class B and as such require a large capacitor to supply current of an opposite direction to the transistor that is on and discharges through the speaker. This capacitor has to fill up-this is done via the speaker hence the displacement, thump. when the supply is switched on. It does, however, fill up to half the supply voltage in a fraction of a second and should cause no concern.

For peace of mind this thump can be avoided by placing a similar value capacitor in position from the mid-point to the opposite supply side.-R. King (Beds.).

\section*{Solid State. That letter}

I would like to take the opportunity to answer the query of Mr. W. J. Tomlinson, in the July edition of Practical Wireless, concerning the "solid state".
This term is of a somewhat nebulous nature to define. Strictly speaking, it applies to the group of substances known as solids, i.e., those substances which possess both definite volume and definite shape; in contrast to these are the substances known as fluids, which depend on their surroundings for their shape.

The study of substances in the solid state, a branch of physics, deals with the structure and properties of solids, and may be divided into three approximate groups. Firstly, the structure of solids, i.e., crystallography, the structure of metals, etc.; secondly, the natural phenomena exhibited by solids, i.e., specific heats, thermal and electrical conductivity, intrinsic semiconductivity, magnetic and dielectric properties, etc., and finally, the defects of solids, such as impurity semiconductivity, lattice defects, etc.
From this rather broad division, it will be seen that included are the substances known as semiconductors. It is to this small group of substances that the term "solid state" has been widely, though incorrectly, applied.

Thus, when Mr. Tomlinson notices the words "solid state" on a piece of electronic equipment, he will now realise this. I hope that these few remarks may prove useful in some way to the people concerned, if not to others.-C. J. Gibbins, B.Sc. (Liverpool).

\section*{S.W.L. Cards}

I have just received a most disappointing B.C. verification card from the other side of the world, the result of many nights of patient listening

On the front, there is a photograph of a landscape which could just as well have been in Scotland, New Zealand or the USSR. The station name, along with the usual verification details is given on the back. This is only one of many I have received. Most S.W.L.s like to display their "catches" on the wall next to the receiver, and therefore prefer the station name, along with a simple design and/or station information to be shown on the front.
B.C. stations should bear in mind that the S.W.L. is primarily interested in the station and not in views of their country, which can be found in most school geography books anyway.-R. Mitchell (Glasgow).

\section*{Son of Instant Silence}

Beginning to tire of large super sets and such like, I decided to return to something very simple, as the weather was hot. So I disinterred an old Tungstalite detector and made a crystal set. Result: utter and complete silence. So I then made a one diode and one transistor set. Result: the same only more so. This made me somewhat wild (reasonable-Ed.), so I set to and made up a two diode and two transistor set from a reliable "expert" design. I tried this out on a long garden aerial. Result: an intensified silence, far more so than the other silences had been.
So it's now to hell with transistors with me. I'm going to carry on with a sturdy little two valve set right away. You can always rely on a jolly good little stout two valver. - A. Trowbridge (Middlesex).


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1 amp., \(6,8,10,12,16,18,20,24,30,36,4\) 3 amp., \(0,12 v\). and \(0-18 \mathrm{v} . \ldots, \ldots, 30,40,48,60,35 /-\) AUTO TRANSFORMERS \(\quad 0-115-230\) จ. Input/Outpat
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\(6 / 6 ; 12 \times 8 \mathrm{in} .4 / 6 ; 10 \times 7 \mathrm{in} .3 / 6 ; 8 \times 6 \mathrm{in} .2 / 6 ; 6 \times 4 \mathrm{in} ., 1 / 6\).

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> The author discusses a practical approach to a means of making a "standard" v.h.f. radio cover other segments of v.h.f. such as the aireraft and amateur bands.

FIOR quite a number of years now, requests, from Practical Wireless readers have appeared in these columns asking for various types of v.h.f. receivers covering frequencies outside the standard Band II allocation ( \(88-108 \mathrm{Mc} / \mathrm{s}\) ). Many asked for a fully tuneable superhet type of receiver with a band coverage of say \(60-170 \mathrm{Mc} / \mathrm{s}\), and although this is possible produces many design headaches as far as the switching of the signal frequency circuits is concerned. An admirable solution would be to utilise some form of turret tuner to switch the aerial, mixer and oscillator tuned circuits and feed the output into a standard \(10.7 \mathrm{Mc} / \mathrm{s}\) i.f./a.f. amplifier. Attempts were made to obtain v.h.f. coverage with a standard television receiver by altering the inductances of the biscuit tuning coils, employed in the turret tuner, and results obtained have been quite resonable only to be spoilt by the extremely wide sound channel i.f. bandwidth which can be up to \(200 \mathrm{kc} / \mathrm{s}\) wide. Consequently when tuning over a particular band of frequencies it was possible to receive two, or sometimes even three, stations at the same time at one particular frequency setting. For the reason stated above this approach was abandoned and other ideas were looked for.

\section*{FRESH APPROACH}

Since the introduction, by the BBC, of v.h.f. local radio stations the v.h.f. Band II has been introduced to many more transistor portables and there has been quite a drop in the price of such items. On acquiring one of these sets, having medium wave, long wave and v.h.f./f.m. coverage, ideas on how to

\section*{Capacitor \(=-\mathrm{O}\)
terminals -O}

Fig. 1 (above): Method of obtaining a smalf capacitive effect by twisting two insulated wires together.

Fig. 2 (below): Alternative trimmer layouts discussed in the text.

(a)

(b)
modify the set for other v.h.f. frequencies were sought after. The standard v.h.f. coverage was 88\(108 \mathrm{Mc} / \mathrm{s}\) and it was soon realised that by the addition of suitable values of capacity, wired in parallel with the aerial and oscillator tuning capacitors, the tuning could be brought down to cover the \(80 \mathrm{Mc} / \mathrm{s}\) commercial band. The first thought was to utilise small trimmers, having a maximum capacity of about \(20-30 \mathrm{pF}\), but the set was so compact and neatly built that there was not enough space for their inclusion.

\section*{WIRE CAPACITOR}

A well known method of obtaining small values of capacity is by twisting together two lengths of insulated wire, the basic idea being shown in Fig. 1. The wire used can be standard insulated solid core tinned copper wire, adjustment of the capacity value being achieved by altering the length of the twisted wires. Two of these capacitors were prepared, each having a length of about 8 inches, and wired across the appropriate aerial and oscillator tuning capacitors. In order to discover which terminals on the tuning capacitor to use, the following procedure should be carried out. Tune to the local Radio 2 station (about \(90 \mathrm{Mc} / \mathrm{s}\) on the tuning dial) and, in turn, adjust each small trimmer screw on the back of the tuning capacitor. Two of these trimmers will control the v.h.f. coverage and the remaining two the m.w./l.w. coverage. If no difference is noted in the Radio 2 signal, on adjustment of a particular trimmer, then the trimmer should be returned to its original position. If, however, the signal decreases slowly in strength then this will be the v.h.f. aerial trimmer, and should be noted as such. If, on the other hand, the signal disappears completely on turning a particular trimmer only fractionally, then this will be the v.h.f. oscillator trimmer.

\section*{TRIMMER LAYOUTS}

Figure 2a shows the four trimmers as mounted on the tuning capacitor and the appropriate terminals to use. Figure 2 b shows another type of tuner layout very often used. If Tcl is found to be the aerial trimmer then the capacitor should be wired across terminals 1 and 2. Similarly, if Tc4 was found to be the v.h.f. oscillator trimmer, then the remaining capacitor should be wired between terminals 2 and 3. The oscillator capacitor was next cut down to

\section*{THE BROADCAST BANDS}

WELL here we are in the autumn/spring transmission period which started on Sept. 1 and finishes on Nov. 3. During this period the conditions on 25 and \(21 \mathrm{Mc} / \mathrm{s}\) provide excellent DX during daylight and early evening hours, with again good signals from the Americas all hours of the day. So now here are the propagation conditions for the main circuits in the United Kingdom.

South Africa: 0800-1400 25, 21 and \(17 \mathrm{Mc} / \mathrm{s}\); 1400\(160025,21,17\) and \(15 \mathrm{Mc} / \mathrm{s} ; 1600-180025,21,17,15,11\) and \(9 \mathrm{Mc} / \mathrm{s} ; 1800-200021,17,15,11,9,7\) and \(6 \mathrm{Mc} / \mathrm{s}\); \(2000-240017,15,11,9,7,6\) and \(5 \mathrm{Mc} / \mathrm{s} ; 2400-020015\), \(11,9,7,6\) and \(5 \mathrm{Mc} / \mathrm{s} ; 0200-040011,9,7\) and \(6 \mathrm{Mc} / \mathrm{s}\); 0400-0600 15, 11, 9 and \(7 \mathrm{Mc} / \mathrm{s} ; 0600-080021,17\) and \(15 \mathrm{Mc} / \mathrm{s}\).

East Africa: 0800-1400 25, 21, 17 and \(15 \mathrm{Mc} / \mathrm{s}\); \(1400-160025,21,17,15\) and \(11 \mathrm{Mc} / \mathrm{s} ; 1600-180025\), \(21,17,15,11\) and \(9 \mathrm{Mc} / \mathrm{s} ; 1800-200021,17,15,11,9,7\) and \(6 \mathrm{Mc} / \mathrm{s} ; 2000-220017,15,11,9,7,6\) and \(5 \mathrm{Mc} / \mathrm{s}\); \(2200-240015,11,9,7,6\) and \(5 \mathrm{Mc} / \mathrm{s} ; 2400-020011,9,7\), 6 and \(5 \mathrm{Mc} / \mathrm{s} ; 0200-0400 \mathrm{11}, 9\) and \(7 \mathrm{Mc} / \mathrm{s} ; 0400-0600\) 15,11 and \(9 \mathrm{Mc} / \mathrm{s} ; 0600-080025,21,17,15\) and \(11 \mathrm{Mc} / \mathrm{s}\).

South Asia: 0800-1200 25, 21, 17 and \(15 \mathrm{Mc} / \mathrm{s} ; 1200-\) \(140025,21,17,15\) and \(11 \mathrm{Mc} / \mathrm{s} ; 1400-160021,17,15,11\), 9 and \(7 \mathrm{Mc} / \mathrm{s} ; 1600-180017,15,11,9,7,6,5\) and \(4 \mathrm{Mc} / \mathrm{s}\); 1800-2000 15, 11, 9, 7, 6, 5 and \(4 \mathrm{Mc} / \mathrm{s} ; 2000-240011\), \(9,7,6,5\) and \(4 \mathrm{Mc} / \mathrm{s} ; 2400-02009,7,6,5\) and \(4 \mathrm{Mc} / \mathrm{s}\) : \(0200-040011,9,7\) and \(6 \mathrm{Mc} / \mathrm{s} ; 0400-060015,11\) and \(9 \mathrm{Mc} / \mathrm{s}, 0600-080021,17,15\) and \(11 \mathrm{Mc} / \mathrm{s}\).

North East Asia: 0800-1200 21, 17 and \(15 \mathrm{Mc} / \mathrm{s}\); 1200-1400 17, 15 and \(11 \mathrm{Mc} / \mathrm{s} ; 1400-160015\) and \(11 \mathrm{Mc} / \mathrm{s}\); \(1600-2200 \quad 11\) and \(9 \mathrm{Mc} / \mathrm{s} ; 2200-2400 \quad 11 \mathrm{Mc} / \mathrm{s}\) only; 2400-0400 circuit closed; 0400-0600 15 and \(11 \mathrm{Mc} / \mathrm{s}\); \(0600-080017\) and \(15 \mathrm{Mc} / \mathrm{s}\).

Australia via Asia: 0800-1000 25 and \(21 \mathrm{Mc} / \mathrm{s} ; 1000-\) 120021 and \(17 \mathrm{Mc} / \mathrm{s} ; 1200-140017\) and \(15 \mathrm{Mc} / \mathrm{s} ; 1400-\) 160017,15 and \(11 \mathrm{Mc} / \mathrm{s} ; 1600-1800 \mathrm{l} 5,11,9\) and \(7 \mathrm{Mc} / \mathrm{s}\); \(1800-200011,9,7\) and \(6 \mathrm{Mc} / \mathrm{s} ; 2000-220011,9\) and \(7 \mathrm{Mc} / \mathrm{s} ; 2200-240011 \mathrm{Mc} / \mathrm{s}\) only; 2400-0600 circuit closed; \(0600-080021 \mathrm{Mc} / \mathrm{s}\) only.

West Coast South America (North of Chile): 1200180025 and \(21 \mathrm{Mc} / \mathrm{s}\); \(1800-200025,21\) and \(17 \mathrm{Mc} / \mathrm{s}\); \(2000-220021,17\) and \(15 \mathrm{Mc} / \mathrm{s} ; 2200-240017,15\) and \(11 \mathrm{Mc} / \mathrm{s} ; 2400-0200 \mathrm{15}, 11\) and \(9 \mathrm{Mc} / \mathrm{s} ; 0200-080011\) and \(9 \mathrm{Mc} / \mathrm{s} ; 0800-100017,15\) and \(11 \mathrm{Mc} / \mathrm{s} ; 1000-1200\) 17 and \(15 \mathrm{Mc} / \mathrm{s}\).

During the last few weeks I have had letters from beginners to DX-ing asking me to list all the DX programmes that they can listen to, so here goes, The programmes listed here are only the ones beamed to Europe, I have put in the frequencies if they are given for the period Sept-Nov., otherwise I have not listed the frequencies.

Sundays: DX-ers calling, R. Australia 0730-0740 on 11,710, 9,560; World Radio Club, \(B B C\), London \(0930-0945\) try \(21 \mathrm{Mc} / \mathrm{s}\) and 15,070 ; DX Window, R. Denmark, Copenhagen 1015-1035 on 9,520; DX-ing

\section*{by CHRISTOPHER DANPURE}

Worldwide, R. New York Worldwide 1930-1935 on 17 and 15; Finland's DX-Club Programme, Helsinki over Finnish Broadcasting Co. 1615-1630 on 15,185.

Mondays: World Radio Club, \(B B C\), London 02450300 on 6,110; Swiss S.W. Merry Go Round, Berne \(0730-0800\) on \(9,535,6,165\) and 3,985; Deutsche Welle DX-Programme 0915-0930 every 2nd Monday on 6,075; Swiss SW Merry Go Round, Berne 1200-1230 on 11,865 and 9,\(665 ;\). Berlin International DX-Club during 1730-1800, 2015-2045, 2200-2230 and 2300-2330; Deutsche Welle DX-Programme 1830-1845 every 2nd Monday on 6,075 ; Swiss S.W. Merry Go Round, Berne \(2000-2030\) on 9,665 and 11,865 or 6,015 ; Emissora Nacional, DX Club, Lisbon during 20452130 on 7,130 and 6,025; R. Stn. HCJB, Quito,DX-party line 2100-2130 on 17,880 and 15,325.

Tuesdays: Sweden Calling DX-ers, Stockholm 11201130 on 9,625 ; Polish Radio DX-programme every 1st Tuesday during \(1830-1857\) on 11,815 and 7,125 ; \(R\). Budapest DX-programme during 2130-2230 on \(11,910,9,833,7,220,7100\) and 3,995 ; Polish Radio DX-programme every 1st Tuesday during 2130-2155 on 11,815 and 7,125 ; Sweden calling DX-ers, Stockholm 2105-2115 on 6,065.

Wednesdays: \(R\). Stn \(H C J B\), Quito DX-party line 0930-1000 on 15,325; R. Prague DX-programme during 1200-1230 on \(15,285,11,960\) and 9,\(560 ; 1630-\) 1700 on 7,345 and \(5,930,1900-1930\) on 7,345 and 5,930 ; R. South Africa DX-corner 1925-1935 on 17,790 and 15,245; R. Bucharest DX-programme during 19302030 and 2200-2300.

Thursdays: R. Nederland DX-jukebox 1442-1512 on 6,020 ; \(B B C\) World Radio Club \(1245-1300\) on \(21 \mathrm{Mc} / \mathrm{s}\) and 15,070; R. Kiev DX-club during 1900-1930 on 11 and \(9 \mathrm{Mc} / \mathrm{s} ; ~ R\). Nederland DX-jukebox 1912-1942 on 6,020, 2012-2042 on 11,730 and 6,020.

Fridays: R. Prague DX-programme every 2nd and 4th during 0700-0755 on 9,575 and 6,055; Finland's Dx-Club programme over R. Finland 1815-1830 on 15,185; R. Bucharest DX-programme during 19302030 and 2200-2300; Trans-World Radio "DX-special" via Bonaire \(2100-2115\) on \(15,245 \mathrm{Mc} / \mathrm{s}\); R. Sofia DXprogramme during 1930-2000 and 2130-2200 on 9,660 and 6,070 .

Saturdays: DX-special. Trans-World Radio, Monte Carlo 0610-0625 on 41,18; Radio Canada S.W. club, Montreal 0730-0740 on 9,625 and 5,990; Radio Japan DX-corner, Tokio 0825-0830 every 1st and 3rd on 21,535 and 17,825 ; Radio Canada S.W. club, Montreal 1235-1245 on 17,820; DXing Worldwide, \(R\) New York Worldwide 1730-1800 on 21,525 and 17,845; R Budapest DX-programme, during 2130-2230 on 11,910, 9,833, 7,220, 7,100 and 3,995; Radio Canada S.W. Club, Montreal 2123-2133 on 17,820, 15,320 and 11,720.

Deadline is the 15 th September, so until next month 73's and good DX-ing.

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WHATEVER happened to poor old 10 metres? I've listened, you've listened, but no one seemed to hear very much at all. The band seems to be getting worse and the openings are rarer and shorter. Anyone offer any theories as to why this should be so just at the time when 10 should be giving us some FB DX? It certainly can't be the gear or the individual since there is such a variety of receivers in use and a vast number of s.w.l's.

Most logs were the result of "easy" DX heard on 20 and 15 metres. Few ventured lower than 20 , and only two logs arrived for 7 megs. How about only listening on 160-80-40 metres this month-go on, I dare you.
J. Baker (Lancs.) says that the 4A callsigns have been given to Mexican stations (normally XE) to commemorate the Olympic Games and may only be used for the period March 31st-December 31st 1968. The SK callsigns are issued only to Swedish club stations and SK6AB was heard to say that there are about 15 SK calls at present. Another callsign confusion comes from Indonesia, where there are now three sets-YB ( 500 watts), YC ( 75 watts) and YD (10 watts). Only the YB stations are permitted to work foreign stations. PK8's are legal but have to apply for new licences within the next year.
W. Mantovani (Yorks.) writes in with his solutions to queries raised in past pages. First, QQ7A claiming to be on Ganzo Istand is definitely flying the Jolly Roger. I \(\varnothing\) ARI is apparently a goodie. The correct call should read I \(\varnothing\) ART and was heard April/May when it was the call of a special station in Florence at some obscure festival. Also reported was the arrival on topband of AC4AN (Tibet) on s.s.b. on July 2nd. The last AC4 I can recall which was heard on topband was found to be located in the Bermondsey area of London! Never mind, there's no cause for Alahma.

I hear that some of the gang from Leyton intend to set up a station at Harringay dog track! I've heard of DX hounds but this is ridiculous. Apparently they discovered that there is a nice fat pipe (or something conductive) which goes right round the entire track. With this as an earth, plus a good vertical, they hope to work some real DX. They also have their sights set on Fred (G3SVK) when he does his Channel Islands DX pedition.

\section*{LOW LOGS}

One interesting suggestion this month is that local nets should take place on 40 metres with a 10 -second pause between overs for any stations from further afield who would like to call in. This would keep the band inhabited and dissuade those nasty commercials from invading too. On the other hand, 40 metres is only \(100 \mathrm{kc} / \mathrm{s}\) wide for \(G\) stations. Anyone any views?
M. Pasek (Notts), QP1 66 into an HRO, 2-element fixed wire beam (inverted V's), hooked some FB DX on 40 metres s.s.b., including-HP1JC, PY2ENR, VK2ABZ, VK2AVA, VK2SA, VK3AHT, VK3HW (at 5 and 9 plus 10 dB ), VK3OZ, VK3ZL, YV1BI, YV1EL.
W. Mantovani (Yorks.), Ham-1, 600ft. long wire (cor, that's what I call a long wire), a.t.u. reports

PA \(\varnothing\) BRM and EI2BG using s.s.b. on 160. On 80 s.s.b. the log reads - EI8BT/P, G3WMZ/5A, IZ6KDB (Ponza Island), WIDRS, W1EBC. Forty metres s.s.b. produced sigs from-CN8BV, DU2BU. EA4JV, EA6BG, LA2PH/MM (near Capetown), PY6NG, PY6WA, PY7ASM, PY7GAY, TA2BK, TJIAL, TU2AK, UW3FA, VK3HW, YV1PW, YV2VO, 5A2MJC, and a 6OIAT who was suspected of being a pirate.

\section*{HIGH LOGS}

Now we pass on to the lush DX pastures of 20 and 15 metres. Certainly no shortage of activity here.
M. Collins (Leeds), HE5O plus a 90 ft end fed snuffed these out on 15 s.s.b.-CE1HU, CE2VX, CN8BV, CN8FV, CO2FM, CP6HI, CR6GM, EL3C, EP3AM, HC5BZ, HK4BFF, HK4BIW, K \(\varnothing F B L\), K6EVR, KC4AM, KC4CA, KP4FA, KZ5BU, LU2OF, LXIRB, OA4OA, OA4ZI, OD5BZ, OX3DM, PJ5BE, PZIBX, SVØWL, VEØAE, W7HRH, YV3KW, YV5CR, ZC4RB, ZD7KH, ZD8CC, ZD8NK, ZD8HAL, ZP3TW, ZS4AA, ZS6AD, 4U1ITU, 4Z4HE, 5H3KJ, 6W8CZ, 7Q7BN, 9G1GD, 9H1K, 9Q5DG, 9U5IV, 9X5AA.
G. Coomber (Essex), HRO-MX, dipole, 15 s.s.b. -CR6BF, EL2AK, ET3REL, HK5MO, JAIQWT, JA3APL, KP4AST, MP4MBB, LU9DM, PY2ARS, SV1BK, YV4QG, ZD8AB, ZP5JB, ZS2GF, 4Z4HF, 9G1FL, 9M2DQ, 9U5CR, 9V1OC.
G. Maitland (Isle of Wight), R107T, PCR30, 50 ft end fed, logged these on 20 metres s.s.b.-CN8AW, CR6FC, CR7IC, EA8AV, EL2Z, EP2JP, FR7ZC, HSIHI, JAICEU, K6JN/P, K6TXQ/MM, KA3TZN, KR6BD, LU2BU, MP4BEU, MP4TCE, TA2EL, TN8BRW, VK3AHF, VQ8AS, G5PP/P/W2, W6BMG, XW8AX, ZD8CC, ZL2EM, ZL3UY, 5V2TS, 9AIU/M, 9G1GD, 9K2BJ, 9M2BD.
A. Robnett (Herts.), CR7OA, PR30, 66ft. end fed, \(14 \mathrm{Mc} / \mathrm{s}\) s.s.b. - CR6DU, HSIMAG, HV3SJ, JAØADY, JX1BH, OA6MI, PY3BXW, SV1CB, TA2BK, VE1ASY, ZE5JU, 5H3JL, 9M2YC.
P. Leybourne (Glos.), HRO-M and Racal RA245, dipole, 20 s.s.b.-HB1AB, HC2CB, K \(\varnothing\) TXF/MM, KP4DAC, KZ5AA, LU6MJ, LU8KAE, MP4TCE, OA6RP, PY6NX, SVØWMM, TA2BK, UO6GR, VE2YA, VE3FIE, VK3MO, VK7RX, VP7DL, W5ZPD, XE1EW, YN1GBH, ZL4BO, 4A1MZ, 6Y5DW, 8R1F.
R. Dinning (Ayrshire), HA-350, dipole, logged these on \(14 \mathrm{Mc} / \mathrm{s}\) s.s.b.-AP6GGB, CP5DB, CR6GQ, FO8BY, FR7ZL, GC2LU, HV3SJ, KL7FBO, KV4FA, OH2AM/P/OHØ, PK1TH, TA2BK, UF6CR, UJ8AC, VK2SB, ZC4RB, 4A3AF, 4U1ITU, 5AITK, 5H3JW, 5H4TH, 9Q5CR, 9Q5HS, plus countless W's.

\section*{CONTESTS}

Unfortunately, the contests list for this month (September) was given last month, so you'll have to dig out your previous copy of P.W.-Sorry.

Please note that all logs should reach me by the 20th of each month. Those arriving later, no matter how good, are, unfortunately, just too late for publication in the current issue, and will be too out of date for the following one.


\section*{AN AUTOMATIC PARKING LIGHT SWITCH}

THIS unit was designed and built to be used on a motorcycle having a small six volt battery, but it can also be used with a little modification on cars.

It would be convenient if the parking lights would switch on automatically, instead of somebody having to go out to turn the lights on when they are needed. If the parking lights are switched on when the vehicle is originally parked, there will be an unnecessary drain from the battery which may lead to difficulty in starting the engine or even ruining the battery due to excessive discharge. This unit was needed in particular for the author's motorcycle because the accumulator is rather small (14 A.H.) and there is no room for a larger battery.
As soon as it gets dark enough to need the parking lights, the automatic parking light switch will turn them on, and will not turn them off until daylight. The device is not affected by the headlamps of moving vehicles if a little sense is used when positioning the photoelectric cell.

\section*{CIRCUITRY}

The light sensitive element is an ORP12 photoconductive cell-an ORP12 was used because the author had one conveniently at hand, and other cells might be cheaper and more effective, but the author has not tried any other types specifically intended for the purpose. However, an OC70 was tried for reasons of economy. The OC70 (or OC71 or any other transistor with a glass encapsulation) is prepared by scraping off the black paint and carefully filing a hole in the glass. The opaque jelly is then dissolved out using carbon tetrachloride (which incidentally is very useful for cleaning and degreasing variable condensers, resistors, switches


Fig. 1: Circuit diagram of the unit as built and tested by the author.
etc) and the hole in the glass covered up to prevent contamination (covered by Sellotape). Like all semiconductors, these modified transistors are sensitive to light, but unfortunately the resistance in the dark (of the one OC70 tried) was too low, causing the lights to be permanently off, and also the sensitivity to light was inadequate for this particular application.

\section*{CONSIDERATIONS}

A simple d.c. amplifier was first tried, but as expected, it was rather inelegant because the lights turned on slowly, causing rather excessive dissipation in the final stages of the amplifier and hence being perhaps rather apt to failure unless large transistors were used. Therefore it was decided to use a Schmitt trigger, which is in effect a d.c. amplifier with infinite gain, and therefore permanently overloaded in either of two states, viz Tr 2 conducting (i.e. "on") or Trl conducting (i.e. "off").

This circuit solved the problem very neatly, the lights now being turned on suddenly when the ORP 12 is dark enough. The consumption when the circuit is "off" is of the order to 5 mA (depending on the amount of light), so that the drain on the battery is negligible, when the lights are off. The switch alone takes about 200 mA (i.e. slightly more than the base current of the OC35) when "on" but this current is small compared with the current taken by the parking lights ( \(2 \frac{1}{2} \mathrm{~A}\) in the author's case).

\section*{OPERATION}

The operation of the circuit is quite simple. Assume first that there is enough light on the ORP12 to keep the lights switched off. The ORP12 has a low resistance, causing increased base current in Trl and hence Tr 1 is conducting heavily with a low voltage between its collector and emitter. This voltage is in fact too low to allow \(\operatorname{Tr} 2\) to conduct, and therefore Tr 3 and Tr 4 cannot conduct either, since they have no base currents. (The effect of leakage is negligible). If the light on the ORP12 now gradually decreases, its resistance will rise until eventually the current through it is insufficient to maintain the collector current in Tr 1 . The collector current of Tr 1 then starts to fall, causing the baseemitter voltage of \(\operatorname{Tr} 2\) to rise and eventually Tr 2 starts to conduct, making the common emitter voltage rise and hence reducing further the current in \(\mathrm{Tr} 1 . \mathrm{Tr} 2\) is now driven rapidly into full conduction and the current through it is multiplied by Tr 3 and Tr4 to supply current to the lights. The \(47 \mathrm{k} \Omega\)

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\section*{ES.10-15} LOUDSPEAKER

As described in ' Wireless World August 1968


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resistor shown dotted was not needed in the original, but it may be necessary with some transistor types (particularly if Tr and Tr 2 are germanium). However, \(\operatorname{Tr} 1\) and \(\operatorname{Tr} 2\) were bought as unmarked, untested silicon transistors and were not selected, but they worked well. Two 2N2926's were at hand and these planar transistors were also tried with no apparent change in performance, as was expected because the Schmitt trigger circuit is very tolerant of transistor parameters. An OC35 was used only because it was the cheapest power transistor that the author could find. Its ratings are quite ample for this circuit (maximum collector voltage 32 V , at Ie \(=6 \mathrm{~A}\), max. collector current 8 A , max. dissipation 30 W ) and since it only dissipates about 1 W it does not need a heat sink. In fact, in the prototype it was soldered directly to the Veroboard on which the circuit was built, the whole lot being wrapped up in \(\frac{1}{4}\) in. thick foam rubber and insulating tape (of the thick black variety) to protect the unit from vibration.

\section*{COMPONENTS}

The component values are not critical. The two \(56 \Omega\) resistors connected in parallel are to act as a \(28 \Omega\) IW resistor, as no 1 W resistors were available in the junk box when this unit was built. The value of the resistor used in this position should be about \(30 \Omega\), but definitely not less than \(25 \Omega\) or the ratings of the OC81 may be exceeded. Since the base current of the OC35 when turned on is about 200 mA this unit will only switch a current of up to 4 A using a typical OC35, though OC35s with a larger \(\beta\) than normal will naturally handle a greater current.

If the unit is used on a car (with a 12 V battery), then the resistor R 2 should be increased to \(1 \cdot 2 \mathrm{~K} \Omega\) with possibly some increase in the values of the other resistors. A relay will have to be used instead
of the OC35. The relay is connected instead of the two \(56 \Omega\) resistors and the OC35 is also omitted. This is because there must be a very good connection from the battery positive pole to "earth" because of the very large current taken by the starter motor.

\section*{PROTECTION}

It is essential that the OC81 is protected when the power is switched off. As the current in the relay winding falls, a large voltage is developed across it and this would almost certainly damage the OC81 if this high-voltage pulse were not suppressed. This if done by means of a diode capable of carrying at least five times the relay current. This diode is connected across the relay windings so that it is reverse biased when the relay is energised.

As mentioned above the unit was built on a piece of Veroboard. This measured \(2 \frac{1}{2}\) by \(1 \frac{1}{2} \mathrm{in}\). with holes 0.15 in . apart. In the prototype the ORP 12 was fitted on the board, but it was later decided to have it separate from the board to make it easier to put it in the best place. The ORP12 was also wrapped up in foam rubber and insulating tape, leaving the "element" exposed and about a foot of thin twin flex was used to connect the cell to the rest of the circuit.

\section*{CONCLUSION}

It may be necessary to connect a resistor in parallel with the ORP12 in order that the lights go on at the right time. A few hundred \(\mathrm{k} \Omega\) would be a suitable value to try first.

The author's unit has easily repaid for the cost and time involved in building it and it has performed reliably since it was fitted to the motorcycle, in spite of the vibration and temperature.

\section*{MODIFYING V.H.F. PORTABLES}
—continued from page 425
\(6 \frac{1}{2}\) inches in length. when it was possible to adjust the associated trimmer so that the Radio 2 transmission was just off the scale at the high frequency end of the band. Tuning back over the dial should now produce several of the higher powered or local commercial signals and one selected at the centre of the scale.

\section*{FINAL ADJUSTMENTS}

The aerial capacitor was then cut to a length of \(7 \frac{1}{2}\) inches and by adjustment of the aerial trimmer it was possible to peak the selected station for maximum volume. Once these adjustments have been carried out the two twisted wire capacitors should be placed parallel with the ferrite rod aerial along the receivers length and bent if necessary to facilitate the fitting of the receiver's back cover. Final adjustments to the trimmers can now be carried out for correct band coverage and maximum volume and the back cover replaced.

The coverage of the receiver will now be from about \(70 \mathrm{Mc} / \mathrm{s}\) to just below \(90 \mathrm{Mc} / \mathrm{s}\), the \(80 \mathrm{Mc} / \mathrm{s}\) band taking up about \(\frac{2}{3}\) rds. of the scale length. The number of stations heard will vary from area to area but if you live near a reasonably large town or city there will be no shortage of signals for you
to monitor. If coverage is required below \(70 \mathrm{Mc} / \mathrm{s}\) or so then the addition of extra capacity across the tuning capacitor should make this possible. Initial adjustments could be carried out by temporarily fitting \(3-30 \mathrm{pF}\) beehive trimmers, and then replacing them with miniature fixed capacitors of the appropriate estimated values.

\section*{COVERAGE}

For coverage of frequencies above \(110 \mathrm{Mc} / \mathrm{s}\) or so, it will normally be necessary to remove a turn or two from the aerial and oscillator coils, depending upon the actual coverage required. This is not advisable unless the necessary test equipment is available. One particular receiver modified for reception over these higher frequencies luckily had a 20 pF fixed capacitor wired across the v.h.f. tuning capacitors for normal Band II reception. Removal of these capacitors brought in the \(144 \mathrm{Mc} / \mathrm{s}\) Amateur Band at about mid-scale, the signals received being peaked for maximum strength by means of the v.h.f. aerial trimmer. Unfortunately, the sensitivity over the 2 metre band was rather poor, and allowed reception of only very local amateur transmissions. Matters could be improved, no doubt, by preceeding the receiver with a suitable transistorised 2 metre preamp or the set could be used, without the pre-amp as a \(144 \mathrm{Mc} / \mathrm{s}\) monitor for your own signals.

\title{
CO de GB2LO
}

\section*{SUCCESS OF RSGB FESTIVAL AMATEUR STATION}

EVER seen an amateur station operating from the pavement of a city street? You would have, if you happened to have been passing the Daily Mirror building between July 8 and 20. As part of the City of London Festival, the Radio Society of Great Britain manned a special studio erected outside 33 Holborn by courtesy of the Daily Mirror who provided the facilities and bore the cost of the portable building.
The callsign of GB2LO was selected in honour of the famous 2 LO which broadcast from Savoy Hill in the early 1920s. And another link with the past was the fact that the site of GB2LO was only about 200 yards from Hatton Garden where the pioneer London Wireless Club-the direct forerunner of the RSGB-held its meetings as far back as 1913.

Through the glass panels of the station building, the general public were able to see amateur radio in operation and a P.A. system relayed both sides of QSO's to loudspeakers in the street. A team of about a dozen radio amateurs gave up their time to operate the station in relays and these included G2MI, G2OS, G3IUZ, G3UML, G5AAM and G6RC. During the run of the activity some 1,500 contacts were made with 108 countries, mostly on 'phone for obvious reasons. Schedules were maintained with the ARRL headquarters station W1AW and with

the British Exhibition station VE3LON at London, Ontario. Most contacts were on 15 and 20 m bands.

The station attracted considerable attention and also welcomed amateur radio enthusiast visitors from 30 countries. As a public relations operation the scheme was an outstanding success although it may be questioned whether the array of KW Electronics equipment conveyed the conventional "amateur" radio image! Also, by virtue of the rotatable 2 element cubical quad aerial 200ft. up aloft on top of the Daily Mirror building, coupled with the "exotic" callsign, it was not so much operating an amateur radio station as conducting a non-stop performance! We hope that Mr. General Public did not go home with the impression that amateur radio was as easy as (or simpler than) calling up Auntie on the GPO telephone!



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\title{
Re-activating \\ MERCURY CELLS \\ BY L.B.STOTT
}

THE primary cell, by definition, is one which generates electricity, usually by chemical action, within itself. The secondary cell accepts electricity from an outside source. stores it, and gives it up as required.

It is usually assumed that there is a rigid line between the two, and that the primary cell cannot accept and store electrical energy from an outside source. Most dry cells in use are of the Leclanche type, and because of their reasonable price little attention has been paid to the possibility of prolonging their life by the supply of current from an outside source.

Such prolongation is, however, quite feasible and an article on the subject was published in the April 1961 issue of Practical Wireless. That article is still quite sound, but only refers to Leclanche cells.

\section*{COSTS}

Lively interest in the subject has recently been aroused by the spectacular increase in the price of some mercury cells. These tiny cells, used in behind-the-ear and other miniature deaf aids were at the beginning of the year suddenly increased in price from 1 s .10 d . to 2 s .9 d . each, and this increase bears hardly on the users, many of whom are pensioners and persons living on fixed incomes.

The cells, in good condition, give a useful life of two or three days, and thus cost 1 s . to 1 s .6 d . a day to run. Many are not sold in good condition. and their useful life is accordingly shorter. Any re-activation which is possible is therefore well worth while. Re-activation is in one way a misleading term; the cells must not be allowed to run down very far before they receive attention, and the process is therefore one of keeping them in good condition over a longer period.

There is on the market at least one "charger", but the price is \(£ 118 \mathrm{~s}\). 6 d ., and to do the job effectively and conveniently at least two, and preferably more, "chargers" are required. This rules the commercial article out of the question for most people on grounds of cost.

\section*{D.I.Y.}

Fortunately it is possible to make up simple "chargers" at very small cost, thus making the operation well worth while, as the effective life of the mercury cell can be at least doubled, and with care three or four times the ordinary life can be achieved. Two such "chargers" are described in this article.

It will be observed that the term "charger" has been used with inverted commas. It is, of course, quite incorrect. The correct term would probably be
"re-activators" but for simplicity we will call them boosters.

The mercury cell should have a voltage when new of 1.4 V , and has a reputed life of \(35-40\) milliamphours. The usual discharge rate is one to two milliamps an hour. The Leclanche cell has a voltage of 1.5 V and quite a small cell can stand up to a continuous drain of 5 milliamps. At a discharge rate of one to two milliamps it will maintain its voltage of at least 1.5 for a considerable time, and this gives us the key to the design of our boosters.

\section*{CHARGING}

The commercial "charger" uses two Ever Ready D14 cells (or equivalent) in series and thus has a voltage of 3 ; a 2,000 ohm resistor in series gives a charging rate of about 1.5 milliamps. The makers advise that the cells be used for 5 hours and then re-charged for 7 hours. It will be seen from this that more than one "charger" is required to maintain three cells (a day's supply at 5 hours each) in good condition.

The voltage used is rather high and if the time is substantially exceeded the cell may be damaged.

If a single Leclanche cell is used the time may be exceeded with impunity. Under these conditions a current of approximately 1 mA flows from the Leclanche to the mercury cell. Typical conditions are as follows:
\begin{tabular}{ccc} 
& \begin{tabular}{c} 
At the \\
outset
\end{tabular} & \begin{tabular}{c} 
After \\
7 hours \\
1.48
\end{tabular} \\
\begin{tabular}{l} 
Voltage of Leclanche cell \\
Voltage of partly discharged \\
mercury cell
\end{tabular} & 1.5 & 1.2
\end{tabular}

After 7 hours the voltage of the two in parallel remains steady at about 1.45 volts and the current ceases to flow. On separating the two the voltage of the Leclanche cell quickly returns to 1.5 and the other drops to 1.4 .

\section*{CELL CONDITION}

One important condition must be observed: it is no use trying to re-activate a nearly exhausted cell; the newer the cell the more efficient the result. The best plan is to bring three cells into use and to have two in boosters and one in the deaf aid. Four cells with three boosters is even better. The writer has five boosters in use and cannot remember when last he bought a packet of six mercury cells; it is certainly more than three months ago. The Leclanche cells have been replaced twice; it is important to have them in good condition, and showing as high a voltage as possible.

The nominal voltage of the mercury cell is 1.4 , but it can drop to 1.25 and still give an effective output. It is better not to let it go any lower. These voltages should be measured on a 1,000 o.p.v. meter if possible. The indication will then be the voltage on load.

\section*{CONSTRUCTION}

The author has two "boosters" and they are essentially identical. The circuit of one is shown in Fig. 1 and is the simplest ever published.

The larger of the two boosters is more suitable for use by an inexperienced person and it will be described first.
The body is a wooden bobbin approximately 2 in . long with a central hole \(\frac{9}{16}\) in. in diameter. The bobbin used in the prototype was originally the core of a roll of paper tape, but another source of material could be the core from a reel of copper wire with the cheeks removed. The central hole must be opened out to a diameter of \(\frac{9}{1}\) in. so that the D14 cell slides easily into it.


Figs. 1, 2 and 3 indicate graphically the simple ideas outlined in the main text.

An additional smaller hole about \(\frac{3}{32}\) in. in diameter is drilled parallel with the larger one, and into this is inserted a length of 6BA studding or a long bolt if available; this should be a tight fit in the hole. On the ends of the 6BA studding are bolted two strips of springy brass, one close to the bobbin and the other fitted with spacers at a distance to ensure that the mercury cell is kept in firm contact with the Leclanche cell. The construction is shown in Fig. 2.
The only critical measurements are those specified.

\section*{ALTERNATIVE}

The other booster, although it could not possibly be simpler, is equally efficient, but it requires greater care in use; if the mercury cell is inserted the wrong way round it would be ruined.
It consists of nothing more than a strip of hard brass with the two ends bent at right angles. The original was made from brass strip \(\frac{1}{4}\) in. wide and \(\frac{1}{15}\) in. thick, but these measurements may be varied according to what is available. It is essential, however, that the inside measurement should be exact, and to take a D14 cell with a 657 mercury cell the measurement should be \(2 \frac{1}{8} \mathrm{in}\). If the distance is slightly exceeded matters can be adjusted by soldering a washer on one, or if need be on both, ends. The cells must be inserted in the booster as shown in Fig. 3.
It will be observed that a positive "earth" has been adopted. The outside of the Leclanche cell, being shrouded by the zinc negative pole, a negative "earth" is more usual. The mercury cell is, however, shrouded by the positive pole, and if the frame of the booster were negative, the positive pole of the mercury cell could easily come into contact with it and this would run down the cell.

The Leclanche cell is fixed to the frame by means of Sellotape, care being taken to ensure firm contact between the positive pole and the frame. The mercury cell is inserted between the negative pole of the Leclanche cell and the frame, care being taken to observe the correct polarity.
The voltage of the Leclanche cell should be checked frequently under load conditions, to ensure that the voltage appreciably exceeds \(1 \cdot 4\) volts.

\section*{CONCLUSIONS}

Both types of booster have been given an extensive field trial by elderly people who are probably below average in mechanical aptitude. Although it cannot by any means be claimed that the devices are foolproof no serious difficulties were encountered. Perhaps the consciousness that a mistake would cost 2 s . 9d. was salutory.
These users were all anxious to know how they could be sure that the devices were operating correctly. Provided reasonably fresh "pen cells" were used they could be assured they were not at fault. It was found useful to show them how to test with a 3 V bulb and to note the condition of the cell by the intensity of the glow. If the pen cells were in good condition the users could easily judge the condition of the mercury cells which, if faulty, would be found to be satisfactory for a few minutes and thereafter rapidly fade away. After a further period on charge, to ensure that a mistake had not been made, the mercury cells would be rejected if they became ineffective within a short period.

The pen cells lasted for about a month, and six mercury cells, using each cell in turn for four hours and leaving all the others in boosters, lasted two months, and some cases longer.

\section*{WIRELESS INDEX}

The index to Volume 43 of Practical Wireless is now available from the Post Sales Department, George Newnes Ltd., Tower House, Southampton Street, London, W.C.2.

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\section*{Construction}

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\hline 184 & 4／9 & \(20 \mathrm{P} 314 / 8\) & ［ L04 5／9 & EM80 5／9 & PL83 7／－ & UF41 & \(9 / 8\) \\
\hline 185 & 418 & 20 P 4 18／6 & 10196 7－ & EMSI 6／9 & \(\mathrm{P}^{\text {PR84 }}\) 6／3 & UF80 & \(71-\) \\
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\hline 3U4G & 4／6 & \(30 \mathrm{C17} 12 / 6\) & JAFF42 8／6 & EY86 6／3 & РM84 7／9 & UL44 & \(201-\) \\
\hline 5 V 4 G & 8！－ & \(30 \mathrm{Cl8}\) 9／－ & EB91 2／8 & EZ40 7／8 & \(\begin{array}{ll}\text { PX25 } & 10 / 6\end{array}\) & U 1.84 & \(6 / 6\) \\
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\(13 /-\) & AF115 & \(31-\) \\
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\hline 6V6G & \(3 / 6\) & 1336 rr \(4 / 9\) & ECLs3 9／－ & \({ }^{\text {PCFP66 }} 8 / 9\) & U301 13／6 & AF127 & 3／6 \\
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\hline \multicolumn{2}{|l|}{52 set Sender and Receiver Circuits \(7 / 6\) post free} \\
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\hline C82 & 4- & EF91 & 2/3 & PCr'so & 8/8 & PY8 & \\
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\hline & 4/9 & 88 & 101- & L81 & 7/6 & & \\
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Micro Sonic transistor radios, 4 nickel cadmium batteriea, battery charger, leather case, store soiled, no guaranitee, OK for repair or spares, in original boxes, \(24 / 8\) post paid.
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Garbon film Low noise
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\(+1 W\) & \(5 \cdot 1 \Omega\) & to \(330 \mathrm{k} \Omega\) & Fiz4 & \(1 / 10\) \\
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A vailable in horizontal or vertical mounting \(1 /\)-each. LOW COST VOLUME CONTROLS: \(100 \Omega\) to \(10 \mathrm{M} \Omega\) \(\operatorname{lin} 2 / 3\) each. \(5 \mathrm{k} \Omega\) to \(5 \mathrm{M} \Omega \log 2 / 3\) each.
CERAMICS: \(1000,2200,4700 \mathrm{pF}, 500 \mathrm{~V}\) 5d, 0.005 , \(0 \cdot 01,0 \cdot 02,0.03 \mu \mathrm{~F}\) 50V 5d. POLYSTYRENE 10pF to 820 pF 5d
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5 ft . 8 in. sections as above \(20 /\) - per section. Carr. 3/. each. Nylon guy lines with semi-autnmatio each P. \& P. 2/. esch. swivel base \(30 /\)-. Corr. \(10 /\) Ground spikes \(4 / 6\) each. P. \& P. \(1 / 6\) each.
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Suital)h all fxings and bane focatlons. Botton suilathe all ixings and bane ocatlons. Bottoin
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Covers \(2-8\) Me/s in 2
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Type B. Secs \(9 v\) at 120 amps car
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Type A. To smooth 8 amps D.C.size \(6 \times 5 \times 5\) in. Type B. To smooth 4 arr.
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