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Now available from stock-superb specifi cation: three $\frac{1}{4}$ track stereo/mono heads 3 speeds- $1 \frac{2}{8}, 3 \frac{1}{4}, 7 \frac{1}{3}$ i.p.s.; takes forward and rewind: tape position indicator; panse control; separate record, replay and erase heads--4 tracks; piano key controls interlocked for all functions; stop) start can be remotely controlled; auto, tapeend stop. Heavy duty motor, capacitor star and run, large dynamically balanced fly wheel. Deck finished in grey plastic, size $14 \frac{1}{3} \times 12 \times 6$ in. depth below plinth 4 in . For I10v. 50 e.p.s. Mains operation, AutoGNS. Carriage \& Packing 10/6 extra
LASKY'S PRICE 291
INTERNATIONAL TAPE Famous American Erand-Fully Guaranteed in. Message tape, 150ft
$\begin{array}{cc}26 & 5 \text { sin. Long play, 1200ft. Acetate } \\ 3 & 9 \\ 7 & \text { Sbin. Standard play, 850it. PVC } \\ 7 & 58 \mathrm{in} \text {. Long play, } 1200 \mathrm{it} \text {. Mylar } \\ 10 & 0\end{array}$ in. Message tape, 225 ft .

Message tape, 300 ft .
3in. Triple play, 600 ft . Mylar
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$5 i n . ~ D o u b l e ~ p l a y, ~ 1200 \mathrm{ft}$. Mylar
oin. Long play, 900 it . Acetate.
jin.
bin. Triple play, 1800 ft . Mylar
in. Double play, 1800 ft Mylar
${ }_{5}^{5} \mathrm{sin}$. Triple play, 2400 ft . Mylar
7 in . Standerd play, 1200 ft . Acet
7 in . Standard play, 1200 ft . Sy . 7 in . Standard play, 1200 ft . My
7 in . Long play, 1800 tt . Mylar $\begin{array}{cl}7 \mathrm{in} . & \text { Long play, } 1800 \mathrm{it} \text {. Mylar } \\ 7 \mathrm{in} . & \text { Donble play, } 2400 \mathrm{ft} \text {. Mylar } \\ \text { 7in. } & \text { Long play, } 1800 \mathrm{ft} \text {. Acetate }\end{array}$ 7in. Long play, 1800 ft . Acetate 126
116
150
450
100
126
19
25
75
75
58

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TRANSISTORISED UHF MINIATURE MODEL shielded metal case only $3 \frac{1}{2}$ x $1 \frac{4}{4}, x$ 3in. Fully tunable - complete with two AF139 transistors. LASKY'S PRICE 3916
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363 Tapedeck, HFTTR3 Amplifier and up to 363 Tapedeck, HFFTR3 Amplifier and up to $10 \times$ bnn. speaker to do justice to the resuits. superbly styled, handsomely fish fised, in
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High quality Celestion $10 \times 6$ elliptical speaker (ilustrated). 30/- Carr. 3/6.

MAGNAVOX 363 TAPE TRANSPORTER


Manufactured to precise limits that permit recording and tape playback to the highest standards set by the Musio Industry. Simple reliable design emplops a siugle bigh-duty motor fiast wind on and rapid rewind control, 3 -speed selection with intercontrol, buspeed selection with interpiano key controls. Speeds 11 $1 \frac{1}{\frac{1}{8}, 1 \frac{3}{2}}$ and $7 \frac{1}{2}$ i.p.s. Wow and flutter $0.15 \%$ on $7 \frac{1}{2}$ i.p.s. Max. spool size 7 in. Playing time up to 120 min , per track from $1,200 \mathrm{ft}$. standard tape. Size $43 \pm \frac{\mathrm{x}}{\mathrm{I}}$ Ilin. plus 5xin. below mounting With it track heads. 813.19 .6 . Add 10/- cartiage and insuranee.

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Easily the best complete tape amplifier available to the home builder. Supplied already matehed for the Magnavox 363 to most other mays readily matched

Features include: switched equaliza tion for all speeds (COIR standards at $7 \frac{1}{2}$ i.p.s.) Treble boost incorporated playis for $3,7.5$, and 15 ohms, additional outputs for extension speaker phone monitoring on Record and Hi-Ft playback through existing systems Inputs for Mic. Pick-up, and VHF Radio, Valves: EF86, ECC83, EL84, EM81, ERZ81. Size overall; $11 \times 6 \times 6 \mathrm{in}$. (Panel $13 \frac{5}{\frac{3}{x}} \mathbf{x i n}$.). Power pack on separate chassis size $7 \frac{1}{2} \times 3 \times 4 \frac{\pi}{3} \mathrm{in}$. Amp. Assombled end tested fig ts carriage.
Carrying Case specially designed to take Magnavox 363 ana HF/TR3 unit. uper fy comlra mish, dark grey

0in. $\times$ Gin. elliptical speaker suitable for use with above.

## TAPE PRE-AMPLIFIER

 TYPE 'C'

Specially developed by Mullard Laboratories for use with high quality replay matohed for use with the Magnavox: 363 tapedeck. supplied speatures included ferroxoube pot core inductors for treble equalization, push puil oscillator heorporating ferroxcube transformer, adjustable output for matching to existing high-quality amplifer systems, inputs for Mic., Pick-up, Radio, etc. Totally enclosed in case size $11 \frac{1}{3} \times 6 \frac{1}{2} \mathbf{x}$ 3 年in. high. (Panel $11 \frac{1}{4} \times$ atin.) Power supply of 300 vde at 25 mA , and 6.3 F , at 1.5 A. is on separate subchassis size $6 \frac{1}{4} \times \frac{4}{2} \times 4 \frac{5}{8} i n, ~ h i g h ~ t o ~$ facilitate remote location from tape heads. Pre-amp and power pack Kit of parts £14. Assembled añ tested
£19. 10s. Add $7 / 6$ carriage.

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 with FULL-FI HEAD 4 -speed, plays 10 records, $18 \mathrm{in} ., 10 \mathrm{in}$. or 7 in . at 16, 33, 45 or 78 r.p.m. Intermixes 7 in ., 10 in ., and play position: colour, brown. Dimensions: 121 x $10_{3}^{3} \mathrm{in}$. Space required above baseboard $4 \frac{3}{3} \mathrm{in}$., below baseboard $2 \frac{3}{2} \mathrm{in}$. Fitted with Ful. Fi-turnover crystal head. Soldering $\operatorname{Ir}$
Multi-Meter

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Size $37 \times 2 \frac{7}{7} \times 1 \frac{3}{6} \mathrm{in}$. Meter size $2 \frac{1}{6} \times 1 \frac{3}{8} \mathrm{in}$. Sensitivity 1,000 O.P.V. on both A.C. and D.C. volts. $0-15,0.150,0-1,000$. D.C. current $0-150 \mathrm{~mA}$. Resistance $0-100 \mathrm{k} \Omega$. Complete With test prods, battery and full instructions, 42/6. P. \& P.
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Precision made. Push-button controls. High quality $2 \frac{2}{4} i n$. speaker. Push-pull circuit. Output 400 mW . Frequency response: $200-7,000 \mathrm{k} / \mathrm{s}$. Fast rewind. Up to 1 hour twin track playing sions: 8 in . x $1 \mathrm{ilin} . \mathrm{x} 3 \mathrm{zin}$. Weighs only 7 ib . Takes sions: 8 in. x $11 \mathrm{in} . x 37 \mathrm{in}$. Weighs only 7 lb . Takes plus $7 / 6$. P.

40W FLUORESCENT LIGHT KIT

bi-pin holders, starter and starter holder.

Incorporating GEC Choke size $8 \frac{1}{4} \mathrm{in} . \times 1 \frac{3}{4} \mathrm{in} . x$ $1 \frac{3}{4}$ in., 2 bi-pin holders,
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Kit incorporating GEC choke size Il $\frac{3}{3}$ in.

17/6
P. \& P. $5 / 6$

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Size $2 \frac{1}{2} \mathrm{in}$. x $2 \frac{1}{6}$ in. approx. By famous manufacturer. A.M.-I.F. $470 \mathrm{ke} / \mathrm{s}$, F.M.-I.F. 10.7 $\mathrm{Mc} / \mathrm{s}$. A.M. coverage from $1,620 \mathrm{kc} / \mathrm{s}-525 \mathrm{ke} / \mathrm{s}$,
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Set of three Tape Deck Motors. These are made for 110 v , but suitable auto. transformer is suppilied. Threemotors 39/6, P. \& P. $6 /-$.

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F.M. TUNER

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4 CHANNEL MIXER TM-I.
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all components for
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AUDIOTRINE HFIOOD IOin. I5W
ML-FI Loudspeakers. Heavy cast construction. Dual cone. Smooth trecoil 3 or 15 ohms. Excep- $95.5 p e e c h$
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15 ohms required. formance. Prease
1.5 ohms required. R.S.C. JUNIOR BASS RERELEX CABINET Designedfor above speaker, but suitable for any good quality $8 i n$. or $10 i n$, speaker. Acoustically lined and
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Handsome
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The GLOUCE simime" cabinet, acoustically lined. Size $24 \times 20 \times 6 \frac{1}{2}$. Finished light
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$\star$ Will amplify direct from Tape Fieads
$\star$ Stereo/Mono switch.

* Separate Bass "Lift", and "Cut" and

Neone panel indicator
$\star$ Neon panel indicator.

Output transformers are high-quality sectionally wound to required specification. Output matching for 3 and 15 ohm point wiring diagrams and instructions.
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FULLY TRANSISTORISED $200 / 250 \mathrm{v}$. A.C. Mains Operation. 10 WATTS R.M.s. into 15 ohms. 15 WATYS R.M.S. into $3-4$ ohms.
Maximum instantaneous Peak power output 28 watts PRINTED CIRCUIT CONSTRUCXION. LATESTMULLARD TRANSISTORS. AD149, AD149, 00127Z, 0081Z, 0C44, 0044, 0081Z, 0044, POSITION INPUT SELECTOR SWITCH GtUALSATION to Standard R.I.A.A. and Fi̛ds TAPE MONITORING FACILITIES. SENSITIYTTIES: Magnetic P.U. 4 mV . Crystal or Ceramic P.U. 400 mV . Microphone 4.5 mV .
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 BASS CONTROL: +12 dB to -15 dB at $50 \mathrm{c} / \mathrm{s}$. HARMONIC DISTORTION at 10 Watts
1,000 c.p.S. $0.35 \%$. HUM LEVEI, - 75 dB . VEGATIVE FEEDBACK: 52 AB

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IMPORTANT NOTE flgures are given in R.M.S. Rated output and music or I.H.F.M., otherwise we could obviously quote muoh higher outputs

D DUAL CHANNEL VERSION OF THE SUPER 15. Employing Twin Printed Circuits. Close tolerance Ganged Pots. Matched Components. CROSS-TAEM: 52 dB at $1,000 \mathrm{c}$.p.s.
CONTROLS:5 position InputS Selector, Bass Control, StereolMono Switch, Tape Monitor Switch, Mains Switch. netic P.U. (2) Ceramic or Crystal P.U. (3) Radio/ Aux. (4) Tape Head/Microphone.
Operation of the Input Selector Switch assures appropriate equalisation.
Rigid 18 s.w.g. Chassis. Size approx. 12in. Wide, Attractive rigid Perspox Facia Plate and Matching Spun Silver Knobs. Neon Panel Indicator. Above facilities, etc., except for Ganging and Complete Kit of parts with full constructional details Circuit wired and tested $30 /$ - extra or Kit with printed $1 \mathbf{1} \frac{1}{2}$ Gns. 9 monthly payments of $22 / 6$ (Total e1.3.14.6). Attractive Walnut or Teak fnished cabinet 5 gns. or Deposit $1 \hat{1}$ - and 9 monthly payments $11 /-$ (Total $£ 5.16 .0$. Or unit factory built and tested. complete with cabinet and with our usual 12 months guaran- 181 Gns.
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$250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a} .0-5-6.8 \mathrm{v} .3 \mathrm{a}$ $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a} .0-5-6.3 \mathrm{v} .3 \mathrm{a}$.
$300-0.300 \mathrm{v} .130 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 0-5-6.3 \mathrm{v} .1 \mathrm{a}$ suitable for Mullard 510 Amplitier ...
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## TOPIC OF THE MONTH

## Pilkington Revisited

THE socialogical and cultural findings of the Pilkington Committee on Broadcasting, which delivered its report in 1962, provided a seven-day wonder which was eagerly exploited by the Press, the Industry and do-it-yourself psychiatrists.

The technical problems discussed in the famous report, however, are still with us in part. For example, despite a strong recommendation to start local broadcasting, the Government of the day took no action, nor have subsequent Governments. As the subject is again "under review" we took out our report and waded through the evidence, reports, statements, opinions in this vade-mecum. The Committee came out in favour of local radio stations, organised by the BBC. And we feel inclined to agree.

As the BBC is already organised on both a National and Regional basis, it would be a natural development to purely local broadcasting. And local station managers could always call upon selected BBC programmes to supplement local items. A comprehensive and flexible network as opposed to the likelihood of non-stop juke-box radio.

Stations run by commercial companies are placed in the invidious position of trying to serve local interests and also obligations to advertisers-which, in the long run, is incompatible. To serve local hopes, endeavours and interests there must be as much freedom as possible from external direction. The BBC, financed by licence revenue, would have no obligation to pursue any objective other than that of the public radio service.

There must be an answerable public corporation to assume responsibilities for a service. The BBC, as such, fulfils this requirement. But a corporation made answerable to a multitude of commercial companies would find it impossible to ensure that the major obligation to the public was met-and the essential purpose of the service would be largely frustrated.

Moreover, the BBC estimate that a nationwide system of local v.h.f. stations could be set up at an additional licence fee of only five shillings. This, we feel, would be good value for the money.
W. N. STEVENS, Editor

## NEWS AND COMMENT

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[^3]
## Simple Proximity Detector

With reference to the article in the March issue of Practical Wireless, "A Simple Proximity Detector" I would like to draw attention to page 940 , column 2, paragraph 4, regarding the d.c. setting of this unit.
(1) It is the responsibility of the Police Force to protect your property and if they are informed of your absence they will do so.
(2) The Electricity Authority cannot as far as I am aware, in the majority of cases, cut off the electricity without entering the premises, so your door is likely to be damaged and very insecure.
(3) The nuisance caused by the ringing could cause neighbours to take civil court action against you for interfering with their peace and comfort and may be able to claim damages for the interference.
All of these points could be overcome if the operator leaves his keys with a responsible neighbour who can turn the implement off and possibly reset it. You must also of course inform the local police of the keyholder's name and address. This will save tying up a member of the understaffed police for some considerable time.
D. E. McNair.

Slough,
Buckinghamshire.

## Wake Up Dealers

I am getting weary in my search for a dealer who will take money from me and, in exchange, do at least one of two thingsnamely, align my home-built f.m. tuner and test for me some valves.

Because I have no proof of where the parts for the tuner were bought the best I have been able to achieve so far, and after almost begging for it, was 17 s . 6 d . worth of very dubious alignment. The tuner very little better than before and no explanation. Possibly the construction is at fault, but as it once worked well I doubt it.

And the valves! London is littered with radio shops whose testers are permanently out of order, not available for use at any time other than those set down by unwritten and ever changing rules, or just not available owing to staff shortage. Where valve testing is done it is usually as a favour rather than a service and I hate paying for a favour.

Own up gentlemen. Both these services can only be performed by you and you are entitled to charge at a rate that makes them economical for you. Isn't it obvious to you that the man who does the job willingly and properly-even if he is charging more for it now than beforeis the man we will go to when we are next spending out on a large item of equipment. Barry Fox.

London N.W.3.

# NEWS AND. . 

## BETTER COMMENTARY FOR RACE FANS

Standard Telephones and Cables Limited used two 1,000W audio amplifiers, 300 horn loudspeakers and some thirty miles of wire to provide Brands Hatch motor-racing circuit with a new sound commentary system.

The loudspeakers, pole-mounted around stands and track, take speech from a possible six different commentary spots throughout the circuit, to sixteen acres of open-air spectator area. Another 25 cabinet speakers are used in indoor enclosures.

The new system provides clearer and more powerful sound using equipment transistorised throughout, except for the amplifier power output stages.

## HOSPITAL SOUND SYSTEM

A new hospital nearing completion at High Wycombe, Buckinghamshire, will have one of the most comprehensive communications systems of its kind.
A $£ 30,000$ installation will enable patients to speak to nurses from their beds, as well as providing the usual radio and television sound programmes. The system, by Hadley Telephone and Sound Systems Ltd., of Smethwick, serves 228 beds in thirteen wards.


Telefunken enter the field of casette tape recorders with their "Magnetophon 401". This recorder, shown above, uses casettes conforming to the "DC System International", a system acknowledged throughout the Continent as standard for this type of equipment.

Battery-operated, the Magnetophon 401 half-track recorder provides a total of 90 or 120 minutes playing time, depending on the casette used. The amplifier employs 12 transistors for its 2 W output. Frequency response is $40-10,000 \mathrm{c} / \mathrm{s}$, and signal-to-noise ratio better than 45 dB . The price is 46 guineas.

## LIGHTHOUSES WIRED WITH BICC CABLES

Two lighthouses-one of legendary interest, the other a unique modern structure-have been wired exclusively with cables manufactured by British Insulated Callender's Cables Limited.

The 155 year old, 115 foot Bell Rock lighthouse, built on Inchcape Rock II miles south east of Arbroath by Robert Stevenson, grandfather of Robert Louis, has been electrified after operating hitherto on oil. Its 730,000 candlepower red and white signal has now been replaced by a 3.5 kW 100 volt tungsten filament lamp.

BICC supplied Bell Rock with approximately 1100 yards of $660 / 1100$ volt mineral insulated cables and accessories.
The new lighthouse is on Kish Bank, 9 miles out in Dublin Bay, and is the largest of its type to be built anywhere. It is 117 feet high and has a $2,000,000$ candlepower light giving a double flashing character every 30 seconds.

The light apparatus, rotated by an electric motor once every 60 seconds, comprises eight 120 volt 1000 W filament type projector lamps in the focus of eight 18 inch diameter parabolic reflectors, each flash being made up by two reflectors. The voltage applied to the lamps is reduced to 90 volts to increase their average life to 800 hours.

# ...COMIMENT 

## BRITAIN'S PART IN MOON SHOT

Britain is providing a vital link in the communications network American astronauts will use in speaking to Earth, when the first U.S. spoce-shot heads for the moon.

The Marconi Company is building a satellite communicationsgroundstation which will be erected on Ascension Island in the Atlantic, ready for the Apollo moon-shot which is planned to put men on the moon.

This photograph shows a model of the station. Parts of the supporting gantry are already on the Island and other sections, including the 42 ft . diameter dish aerial and electronic sub-systems are under test at Marconi's Chelmsford factory.


## COMPACT STEREO AMPLIFIER

Silicon transistors are used throughout a new stereophonic amplifier made by Goodmans Industries Limited. Feeding into $8 \Omega$, this amplifier, the Maxamp 30, will deliver 15 W per channel with a total harmonic distortion claimed to be less than $0.4 \%$ (at $1,000 \mathrm{c} / \mathrm{s}$ ).
The Maxamp 30 measures only $10 \frac{1}{2} \mathrm{in}$. $\times 5 \frac{1}{2} \mathrm{in}$. $\times 7 \frac{1}{4} \mathrm{in}$. and its polished wood cabinet contains integrated pre-amplifier and power-pack as well as the amplifier itself. The amplifier features all the usual controls and facilities (including provision for stereo headphone listening) and costs $£ 49 \mathrm{lOs} .0 \mathrm{~d}$. Frequency response is $20 \mathrm{c} / \mathrm{s}-20 \mathrm{kc} / \mathrm{s}$.

DERBY WINNER


A few months ago, Derby and District Amateur Radio Society held their Annual Dinner.

Over 180 people attended the Dinner and members voted it a great success. During the evening Mr. A. G. G. Melville, the Society's President, presented awards won by various members during 1965. Our photograph shows Mr. R. E. F. Street (left), Derby and District Vice-Chairman, receiving the Founder Members' Trophy for winning the Constructors' Contest.

## Don't Sink the Pirates

Whilst one must agree with the sentiments expressed in the Editorial of the April Practical Wireless, I nevertheless feel that our off-shore buccaneers have proved useful in a few respects.

Firstly they have probably indicated to the BBC that there is a healthy demand for "Wallpaper" music throughout the day, and that in a fun-crazy trend-setting 1966 England the younger generation look for "live" radio. Whether they are going to get it after we finally sink the pirates is another matter, but they have a case if our legal system still wishes to live up to its claim of catering for all tastes.

Secondly it seems to have proved that a commercial radio network of some sort or another would go down well over here (possibly via v.h.f. on a local basis), and that leading manufacturers and traders would not be slow in coming forward to take advantage of it. Critics have made much of the possible evils of endless advertising breaks, but I hardly consider this to be a particularly valid argument. If the nation's eyeballs can lap it up on TV without undue ill effect, likewise the ears should be able to take it.

In conclusion, I shouldn't worry too much about interference complaints from Eastern Europe whilst our pirates eke out the threatened last days. I have not noticed any particular eagerness on the part of the Communist bloc to honour frequency agreements, and for a good example of the "Law of the Jungle" what could be better than the new Peking transmitter on approximately $1525 \mathrm{kc} / \mathrm{s}$ ? It nearly blots out Caroline at times ... !
P. H. Dobbs.

Westbury-on-Trym, Bristol.

## No. 19 Set Mods

Your contributor S. Simpson is to be congratulated for his article on No. 19 Set modifications. At no time can I recall a more explicit set of instructions for carrying out a modification.
Unfortunately I do not possess a 19 set but do have an R1155 and BC348 which I would like to modernise. As these sets are quite common, many readers, I am sure, would like to see similar articles on these.

I wonder if there is a volunteer amongst your contributors who may have done similar mods on the R1155?
R. E. Robinson.

Darlington.
Well, how about it ?-Editor.

## More News and Comment on Page 200

# GRID DIP OSCILLATOR <br> AT some time or other most radio constructors find themselves in need of frequency checking apparatus. Because of this, the familiar g.d.o. is frequently found in amateur stations for, although very precise frequency checking is scarcely possible with the device, it is a most useful one and is sometimes considered handier than a signal generator. <br> The usefulness of a g.d.o is dependent on its calibration accuracy and holding stability, therefore in a home-built item care is required, firstly to con- <br>  

struct a physical rugged specimen, secondly to ensure reasonably good calibration and thirdly to obtain an attractive unit.
G.D.O. circuitry and uses is already well-known so the emphasis here is on construction, since converting a circuit diagram into a satisfactory practical physical form is not always easy. The prototype is attractive in appearance and it can, with care, be copied easly. A fair amount of work is involved, but only simple tools are needed; to construct the prototype, for instance, a 3 in . vice, a hand drill and a few files were the only items used to fashion the metal work.

In the prototype, power requirements are met via a separate power supply unit which is also used to power various other items from time to time. Space does exist, however, for an internally fitted power unit where considered necessary.

## Circuitry

Looking at Fig. 1, valve V1 is arranged in an oscillatory type of circuit, coil L and capacitor VC1 forming the main frequency-determining components. At switch-on, the oscillator produces valve grid current and this is recorded by meter M inserted at the earthy end of R1. Adequate sensitivity demands use of a meter of $500 \mu \mathrm{~A}$ f.s.d. or better, VR1 ensuring that at no time can the meter be over-driven.

Fitment of the closed circuit jack socket is beneficial, for headphones may be plugged in for


Fig. I: Circuit of the oscillator.

## A.S. CARPENTER G3TYJ

monitoring purposes, or an audio signal may be injected to modulate the r.f. signal being generated. When no jack plug is inserted the g.d.o. functions normally.

The g.d.o. may also be used as an absorbtion wavemeter if its h.t. supply is disconnected and if this facility is required a simple toggle or slide switch should be inserted at point " $\times$ ". Such a switch may be mounted on the front panel to the left of the indicating meter. Since an excellent absorption wavemeter already exists at the author's location the facility was not necessary.

No calibration of the meter scale is necessary. Calibration scales are associated with VC1 plug-in coils (L) enabling unbroken coverage of the frequency range $1 \cdot 75-150 \mathrm{Mc} / \mathrm{s}$ this embracing virtually all amateur bands. U.H.F. bands are not accommodated, a separate device being recommended at these frequencies.

Capacitor VC1 consists of a $2 \times 75 \mathrm{pF}$ specimen pruned from a discarded RF27 unit, but other types are usable-the Jackson 02 for example, or a suitable split-stator item.

## Constructional

The main casing consists basically of two L-shaped pieces of 16 s.w.g. aluminium-Fig. $2 a$, $b$-section $a$ carrying most of the assembly. The final length of each section is $9 \frac{3}{4} \mathrm{in}$. but it may be beneficial to commence with pieces 12 in . long and 4 in . wide, making the bends as indicated but leaving an oddment to be cut away from both ends of each piece later. In this way neatness is assured and matching sections result.

The front panel and end-plate cut-outs may then be marked out as is shown in Fig. $2 a$. In the absence of more refined tools, a series of small holes should be drilled along the inners of the cut-outs marked, after which the unwanted metal may be carefully pruned away leaving ragged edges which may be cleaned up with a file.

The panel becomes progressively weaker as the work proceeds but this is not too important for rigidity returns with the fitment of components. The

section shown in Fig. $2 b$ is also prepared along the lines indicated.

## The Main Chassis

To avoid defacing the front panel unduly, the bulk of the construction is carried on a small chassis, the scheme being shown in Fig. 3. This chassis is constructed and wired separately, eventually being located and held by a pair of retaining brackets. Control shafts of VC1 and VR1 then pass through the front panel, a $2 \frac{3}{4} \mathrm{in}$. diameter drum of the type used with cord drive tuning mechanisms being first fitted to VC1.

A piece of stiff white card on which arcs are drawn in Indian ink is glued to the drum flat surface. The card measures 3 in. in diameter and carries the calibration. A piece of perspex affixed to the front panel affords protection and keeps out dust; a cursor line is scribed and inked in.

Details relating to the main chassis and which completes the metalwork are shown in Fig. 3 and are self-explanatory.

The valve holder may then be fitted as indicated and this assembly wired as far as is convenient. The meter may then be mounted on the front panel casing together with the international valve holder and the jack socket. The main chassis is then affixed after which final connections are made using tags 1 and 5 of the octal valveholder to take the connections from VCl. A tag strip bolted under one of the retaining bracket bolts may be used as an anchor point for the 3-core power supply cable from the p.s.u.

It should be noted that at this juncture no power should be applied, or damage to the valve will result!

## Coils

Prototype coils are wound on plastic formers of $1 \frac{1}{2} \mathrm{in}$. outside diameter and 2 in . long, force-fitted on to the bases of discarded octal valves of the 6 K 8 , 6 K 7 , etc., variety, the glass bulbs and internal structures having been removed.

Before smashing the bulb of an unwanted valve it should be placed in a paper bag. Holding the base of the valve firmly the glass bulb is tapped smartly with a hammer! Careful removal of all debris leaves a strong former which, unfortunately, is not long enough for g.d.o. purposes. Plastic, paxolin, or even stiff card, tubing suitable for fitting over the base is now sought and fixed firmly.

At this point a test coil of about 12 turns of


Fig. 4: General layout of main components.
enamelled copper wire should be wound up and plugged in to the g.d.o. Power is then applied and VR1 adjusted to give a meter reading of approximately half scale deflection. If the coil turns are now gripped firmly between a finger and thumb the meter reading should decrease, thus indicating that the device is functioning. The g.d.o. may then be switched off and the sample coil removed for subsequent amendment.
A total of seven coils are needed, plus a loop or hairpin coil, and while details relating to each range are given in Table I, in another construction variations are likely. This is of no importance, the main requirement being to obtain overlapping coverage from range to range.
table I

| Coil | Turns | Spacing | S.W.G.* | Range in <br> $M \mathrm{Mc} / \mathrm{s}$ |
| :---: | :---: | :---: | :---: | :---: |
| LI | 76 | Close | 30 | $1.75-3.50$ |
| L2 | 40 | $"$ | 24 | $2.80-5.50$ |
| L3 | 28 | $"$ | 24 | $5.0-9.0$ |
| L4 | 12 | Wire dia | 24 | $8.50-16.5$ |
| L5 | 5 | $"$ | 24 | $16.0-32.0$ |
| L6 | $2 \frac{1}{2}$ | $"$ | 20 | $31.0-60.0$ |
| L7 | $1 \frac{1}{4}$ | $"$ | 20 | $45 \cdot 0-80.0$ |
| L8 | Loop | - | 20 | $70.0-150.0$ |

*enamelled copper wire.
Note: L8 consists of a hairpin loop $\frac{1}{2}$ in. long wired across the pins I and 5 of a octal valve bases of the type used in metal valves; type $6 \mathrm{H} 6,6 \mathrm{SH} 7$, etc.

The "cut and try" coil winding method adopted was first to wind a former full of 30 s.w.g. enamelled copper wire and then remove turns experimentally using the g.d.o. and a wavemeter which was adjusted to $1.75 \mathrm{Mc} / \mathrm{s}$, the vanes of VCl in the g.d.o. being fully enmeshed. Immediately the signal due to the g.d.o. was detected, the turns left on the coil were counted and these were found to number 76.
This coil was then made L1 and, with the vanes of VCl opened, a check was made with the wavemeter to find the high frequency point. For L2 slightly over half the number of turns used for L 1 were wound on. The wavemeter was then set slightly 1.f. of the highest frequency reading found with L1 and VCl readjusted to full capacitance.

Again a few turns were removed until the wavemeter gave an indication whereupon VCl was reset to the opposite end of its travel to find the high frequency point for the coil. This procedure was adopted until all coils showed overlapping frequency characteristics although as yet no actual calibration had been attempted. Windings were then sealed and doped.
If no wavemeter exists, a communications receiver could be employed or the oddment of circuitry shown in Fig. 5 used in conjunction with a signal generator. Here, socket SK1 is the generator output socket and $L$ is a coil of some eight turns of


Fig. 5: How a signal generator may be utilised for coil checking in conjunction with an additional oddment of circuitry.
enamelled copper wire about 1in. in diameter. A meter with a full scale sensitivity of around 1 mA is connected at the $M$ terminals or the workshop testmeter suitably adjusted may be used.

If the g.d.o. coil is brought close to $L$ the current reading due to the signal generator and seen on the meter connected to terminals $M$ will increase. Immediately the g.d.o. is tuned to the same frequency as the signal generator a violent kick will be indicated by the pointer of the externally connected meter.

## Calibration

Before attempting calibration, the perspex cursor plate and the scale should receive attention along the lines shown in Figs. 6a,b. The perspex cannot be


Fig. $6 a$ (above): The calibrated scale. Note that the low frequency scaling is close to the shaft whilst the outside compartments are reserved for the higher frequencies. This scale must not be taken literally; it is merely a guide, and in any case the rotors of the tuning capacitor used were capable
of a $360^{\circ}$ movement!

Fig. $6 b$ (left): Dimensions of the perspex cursor.

## $\star$ components list

\section*{Resistors: <br> RI 47 k ת <br> R2 $22 \mathrm{k} \Omega$ I watt <br> Capacitors: <br> | Cl | 100 pF silver mica | C 4 | 1000 pF ceramic |
| :--- | :--- | :--- | :--- |
| C 2 | 100 pF silver mica | VCI | $2 \times 75 \mathrm{pF}$ (see text) |
| C 3 | 1000 pF ceramic |  |  | <br> Valve: <br> VI 6C4 <br> Meter: <br> $0-500_{\mu} \mathrm{A}$ miniature plastic-type panel meter. <br> Miscellaneous: <br> Tuning drum $2 \frac{3}{4} \mathrm{in}$. diameter, closed circuit jack socket, B7G valve holder, preferably ceramic, I.O. valve holder, Control knobs (2), 3-core mains type lead, On/off toggle or slide switch- 250 V d.c., oddment perspex, wire for coils, bases for coils (see text), 16 s.w.g. aluminium, paxolin or plastic tubing, etc.}

## Extras to include P.S.U. Item:

Miniature transformer-mains a.c. input. Secondaries: $0-200 \mathrm{~V}$ at $25 \mathrm{~mA}, 6.3 \mathrm{~V}$ at 1 A . Half-wave rectifier, Electrix contact cooled type 250 V d.c. at 50 mA . Miniature tubular electrolytic, $16+16 \mu \mathrm{~F}, 275 \mathrm{~V}$ wkg. One $1500 \Omega$ resistor, 1 watt.
placed in position until calibration has been completed so care must be taken to ensure that the line scribed on it agrees exactly with that drawn on the card scale.

A piece of stiff white card is then placed across the g.d.o. scale cut-outs and fixed with sellotape in such a way that one edge occupies the position later to be taken by the scribed line.

Using a pin-sharp pencil point, calibration marks are made lightly on the scale, the final marks being filled in later in Indian ink with a mapping pen. Calibration up to $30 \mathrm{Mc} / \mathrm{s}$ is easily accomplished using a communications receiver and cross checking with a crystal marker. The signal generator method previously mentioned may also be employed with rather less accuracy perhaps but may be necessary in any case for the highest frequency ranges. Any crystals that are around can also be made use of, as may MSF and other similar transmissions.

## Finalising the Unit

If a self-contained unit is required, the circuitry and components of Fig. 7 may be inserted, these being placed inside the casing beneath the meter. The sides are easily filled in using expanded metal speaker fret, the edges of which are folded to give increased strength. The casing may then be lacquered or spray painted to taste, after which suitable legends may be applied, preferably through -continued on page 189


Fig. 7: A suitable power circuit that can be included if required.


THIS simple short wave receiver has been designed specifically for the young and relatively inexperienced constructor. The author has memories of his own schoolboy experiences, which in retrospect appeared to consist of recurrent financial crisis which more often than not coincided with an item of more than particular interest in these pages! His own interest in amateur radio was sparked off by the description, in an extremely early edition of "Practical Wireless", by an article entitled "The Solo Knob Three". As its name implies, the receiver controls were compounded into a single knob! Although the present day reader may smile at this description, the receiver certainly met with an enthusiastic reception. Unfortunately, due to the then relatively high cost of components, the author's receiver never quite got off the stocks!
Bearing these nostalgic recollections in mind, the author has endeavoured to keep the cost of the receiver described in this article down to an absolute minimum. Although the simplest materials have
been employed, the performance of the finished receiver is extremely pleasing.
The total building cost of the receiver is roughly $£ 4$ to $£ 5$.

## Circuit Description

The circuit diagram of the receiver is given in Fig. 1. For the benefit of the young reader a fuller description than usual is given of the functions of the various components employed in the circuit. V1, which is a 6 SH 7 pentode, functions as a grid leak detector. The coil, L2, in conjunction with the tuning capacitor, VC2, constitutes the tuned circuit. Radio frequency energy from the aerial is fed via VCl to the coupling coil, L 1 , which is inductively coupled to L2.
The purpose of the variable capacitor, VC 1 , in the aerial lead, is to enable the aerial coupling to be varied. At the same time it helps in the elimination of dead spots. These so called dead spots are caused

## THE <br> BY H. WEBSTER

when the aerial absorbs energy from the tuned circuit. When this occurs, regeneration, on which the leaky grid detector is dependent for its sensitivity, is difficult to obtain. Regeneration is obtained by feeding back energy in the correct phase to the grid circuit-an example, incidentally, of positive feedback. The magnitude of this effect depends chiefly on the size of L3 and C2, the proximity of L3 to L2, and the gain of V1. In the receiver the gain of V1 and hence the degree of feedback, is controlled by varying the screen voltage by means of the potentiometer VR1.
After rectification the signal is fed via C6 to the volume control, VR2, and then on to the grid of


Fig. 1: Circuit diagram of complete receiver and power pack.

Fig. 2 (right): Drilling dimensions for chassis (shown folded flat for clarity).
Fig. 3 (below): Drilling dimensions for panel.


V2 which is a medium impedance triode. The amplified signal developed across the V2 anode load resistor, R 7 , is fed to the high resistance headphones via the coupling capacitor, C8.

Power supplies for the receiver are derived from a simple half wave rectifier circuit. Since the current demand is quite modest, a midget type of mains transformer may be used. Complete wiring details of this unit are given in Fig. 8.

## Construction

The prototype was constructed on a $9 \times 5 \times 3 i n$. universal chassis. The front panel is a nominal 10 x 7 in . chassis top plate from the same chassis range. The main drilling dimensions for the chassis and panel are given in Figs. 2 and 3. Note that the panel drilling details are only applicable to the specified dial. When the valve holders are mounted make sure they are orientated correctly as shown in Figs. 4 and 5 . Similarly ensure that the 6 pin coil holder is mounted with pins 1 and 4 pointing to the rear of the chassis. The correct mounting of VC1 on the front panel is of some importance. It will be seen in Fig. 3 that the mounting hole for VC1 is larger than that required for VC2. This is to enable VC1 to be mounted so that it is insulated from the panel. Two fibre washers are used for this purpose, one on each side of the panel as shown in Fig. 4. If this precaution is neglected, no signals will be obtained, since the aerial input will be effectively shorted to earth via the rotor shaft of VC1.

Although wiring of the receiver is quite straightforward a few tips concerning wiring techniques are given to help the absolute tiro.


For successful soldering a really hot iron is essential and each soldering operation should be conducted as quickly as possible. Whereever possible keep the leads short and stiff to minimise the effects of vibration. Of course don't carry this to extremes and cut the component leads too short, or damage to the component may result during the soldering operation. As an example of the correct technique to adopt, consider the wiring of the 7 way tag strip at the rear of the chassis. It will be seen that R2, R3 and C3 are soldered to one tag as shown in Fig. 6. Do not solder each component separately to the tag, instead thread the three leads through the tag hole and solder all three simultaneously. As a guide, the lead lengths of R2


Fig. 4; Layout of components on top of chassis.
and R3 can conveniently be about $\frac{3}{4}$ to 1 in . Although the method may appear obvious it is surprising how many constructors make a multi soldered joint (no pun intended!) out of such an operation. The general outcome is an unsightly blob of solder. The same remarks apply to all joints where two or more wires are joined to the same tag.

When soldering the electrolytic capacitors C5 and C7 into circuit ensure that the correct polarity is observed. This type of capacitor is generally marked at the positive end, either with a red spot or a + symbol.

The radio frequency choke in the anode circuit of V1 is connected at one end to tag 4 of the V2 valveholder. This tag is merely used as an anchorage point since V2 has no internal connection to pin 4.

## Coil Winding

The receiver covers the h.f. band $2-30 \mathrm{Mc} / \mathrm{s}$ in three ranges. The coils are wound on Eddystone 6 pin formers as shown in Fig. 7. The required number of turns for each range is given in the table. Begin by winding on L2, and in the case of range 1 ensure that sufficient space is left at the top of the
former for the aerial coupling winding L1. When winding the coils, keep the wire reasonably taut on the former so that the resulting coil is rigid. A sloppy winding will result in poor frequency stability. Take care that the reaction winding, L3, is wound in the correct sense as shown in Fig. 7. On ranges 2 and 3 the aerial coupling coil is interwound at the earthy end of $\mathbf{L} 2$.

Before winding any of the coils work out the approximate space occupied by each winding and "then drill the holes in the former so that when the wire is subsequently threaded through the holes it takes the shortest possible path to the pins. Avoid any criss crossing of wires inside the former.

## Power Pack

This unit is built on a $4 \times 3 \times 2 \mathrm{in}$. deep aluminium chassis as shown in Fig. 8. The mains transformer should be an upright mounting type delivering about $200-250$ volts at $20-30 \mathrm{~mA}$ and 6.3 volts at 1 ampere. A small midget choke used in conjunction with C9 and C10 ensures a hum-free d.c. output. The metal rectifier is of the half wave type. Almost any type is suitable, provided that it is rated at 250 volts and is capable of passing a few
 milliamps. The wiring details given in Fig. 8 are self explanatory.

## Testing

When the receiver has been completed a few simple tests should be carried out prior to connecting the receiver to the mains supply.
If the constructor has access to a meter the


Fig. 6: Simple continuity test.
Fig. 5 left: Wiring of receiver, below chassis.

## * components list

```
Resistors:
\begin{tabular}{llrl} 
RI & IM \(\Omega\) & R3, R4 & \(27 \mathrm{k} \Omega\) \\
R2, R5, R7 & \(100 \mathrm{k} \Omega\) & R6 & \(2 \cdot 2 \mathrm{k} \Omega\)
\end{tabular}
Potentiometers:
VRI \(50 \mathrm{k} \Omega\) wire wound
VR2 \(\quad 500 \mathrm{k} \Omega\) carbon, log.
Capacitors:
C1, C2
100 pF silver mica
\(\mathrm{C} 6, \mathrm{C} 8 \quad 0.05 \mu \mathrm{~F}\) tubular 350 V
C3, C4 \(\quad 0.1 \mu \mathrm{~F}\) tubular 350 V
C5 \(\quad 1 \mu \mathrm{~F}\) electrolytic 350V
C7 \(\quad 25 \mu \mathrm{~F}\) electrolytic 25 V
\(\mathrm{VCl} \quad 25 \mathrm{pF}\) Wavemaster
VC2 160pF Wavemaster
Valves:
VI 6SH7,6AC7 V2 6J5,6C5
```


## Miscellaneous:

```
Coil former, plain, type 537, Stratton. Coil formers, threaded, type 538, Stratton (2 required). Coil holder, 6 pin, type 964, Stratton. International octal valve holders (2). Tag strips, 7 way (1), 4 way (1). Aluminium chassis, universal type (Home Radio), \(9 \times 5 \times 3\) in. Aluminium panel, \(10 \times 7 \mathrm{in}\). Dial drive (Jackson 4489). Jack socket (Bulgin). Jack plug (Bulgin). Headphones (high resistance type). R.F. choke, type 737 (Stratton). Transformer, \(200 / 250 \mathrm{~V}, 40 / 50 \mathrm{~mA}, 6.3 \mathrm{~V}\) IA, R.C.S., Croydon. Smoothing choke, \(10 \mathrm{H}, 30 \mathrm{~mA}\). \(8+8 \mu \mathrm{~F}\) electrolytic capacitor (C9, Cl0) 450VW. Metal rectifier, \(250 \mathrm{~V}, 40 / 50 \mathrm{~mA}\). Screws, wire, etc.
```

resistance between HT + and the earth line of the receiver should be measured. It should be roughly $75 \mathrm{k} \Omega$. Also check the heater circuit. The valves must not be fitted for this test as their heater elements are of low resistance and thus, leaks will not show up. If no meter is available a few rough and ready tests may be carried out with a 3.5 volt flashlamp bulb and a 4.5 volt battery as follows.
Connect the lamp and battery in series across the


Fig. 8: Construction of power pack.

mportant,
pins numbered looking down into coll former.

Fig. 7; Method of coil winding and table of data.

| Range | L] | 12 | L3 |
| :---: | :---: | :---: | :---: |
| 2.0/5Mc/s. | $\begin{gathered} 6 \mathrm{t} \\ \text { close wound } \\ * \end{gathered}$ | $\begin{aligned} & 40 \mathrm{t} \\ & \text { close wound } \\ & 26 \text { s.w.g. } \\ & \text { enamelled } \end{aligned}$ | 5t close wound spaced $\frac{3}{16} \mathrm{in}$. from L2 |
| $5.0 / 12.0 \mathrm{Mc} / \mathrm{s}$ | $3 t$ interwound at earthy end of L2 | $\begin{array}{\|c\|} \hline 15 \mathrm{t} \\ \text { wound } 14 \mathrm{t} . \text { p.i. } \\ 18 \text { s.w.g. } \end{array}$ | $1 \frac{1}{2} \mathrm{t}$ close wound spaced $\frac{1}{\mathrm{i}} \mathrm{in}$. from L2 |
| 12/30Mc/s | $2 \frac{1}{2} \mathrm{t}$ interwound at earthy end of L2 | $\begin{array}{\|c\|} \hline 4 \mathrm{t} \\ \text { wound } 14 \mathrm{t} \text {.p.i. } \\ 18 \text { s.w.g. } \end{array}$ | 23 t close wound spaced $\frac{3}{16}$ in. from L2 |

* This coil wound on Eddystone plain former type 537. LI wound at aerial end of L2 and spaced $\frac{1}{8} \mathrm{in}$. from L2.
Remaining coils wound on Eddystone threaded former type 538.
LI and L 3 on all coils wound with 26 s.w.g. enamelled wire.
receiver HT + line and earth as shown in Fig. 6. If all is in order the lamp will remain unlit. Naturally this test will only show up a dead short or a resistance of a few ohms. The same test may also be applied across the heater leads. If the lamp lights up in either of these tests it is imperative that steps are taken to find out the cause of the short circuit.

The power pack may also be tested in a similar manner by connecting the tester across the HT+ and HT - leads. If these tests are satisfactory the receiver may now be put into operation. Plug in the range 2 coil (on which there is a good deal of activity) and the two valves. The headphones are plugged into the jack socket at the front of the panel.

Connect the power leads to the appropriate terminals on the power pack, connect the mains supply, and the valve heaters should light up almost immediately. A slight background noise should also be heard in the phones. Advance VR2 to maximum and then slowly advance the regeneration control from its minimum position until the receiver is just on the verge of oscillation. Tuning may now be done with VC2. The constructor will find that the position of the regeneration control will not remain constant over any given frequency band but will require adjustment from time to time. Always work with the regeneration control set so that the receiver is just on the verge of oscillation. The receiver is then in its most sensitive condition.


## ELECTROSTATIC RECORDING

SINCE every magnetic effect has an electrostatic counterpart, it is rather surprising that electrostatic recording has been so neglected, compared to magnetic recording. Indeed, it is only recently that serious attempts have been made to develop the techniques and to devise a theory of the recording mechanism.

In conventional magnetic recording, we use a magnetic tape and record by modulating the current through an electromagnet, the recording head. In electrostatic recording, we use a dielectric tape and modulate the voltage on an electrode.
In magnetic recording, we can eliminate to a very great extent the effect of the nonlinear shape of the magnetisation characteristic of the tape by high-frequency bias; and a similar effect is found with electrostatic recording.
In both systems, the high-frequency response is dependent on the construction of the recording heads, but the electrostatic system has advantages at the low-frequency end of the scale.
A typical electrostatic recording system is shown in Fig. 1. The tape is drawn between two knife-


Fig. 1: Typical electrostatic recording system.
edges, one earthed and the other connected to the signal and the bias source in series. The bias is of the order of 100 volts at $350 \mathrm{kc} / \mathrm{s}$.
For replay, the tape is again drawn past a knifeedge, this time connected to an amplifier. The output is considerable by magnetic recording standards, some 40 mV for an input resistance of one $\mathrm{M} \Omega$.

Even with such a crude system, and using d.c. bias instead of the a.c. bias system shown, quite reasonable results are obtainable, certainly better than can be obtained with magnetic recording using a permanent magnet bias system. D.C. bias should be about 1 kV , and a circuit of a suitable system is shown in Fig. 2.

The frequency response of this system is rather restricted, however, and for a more ambitious


Fig. 2: Simple electrostatic circuit using d.c. bias.
system, a.c. bias must be used, preferably with the refinement of shielded heads.
The construction of heads for electrostatic recorders is very much easier than the corresponding task for magnetic recorders, and much remains to be discovered about the best way of making such a head.

For the simple system, the best possible electrodes are razor-blades; there are few sharper edges available to anyone. Many of the modern stainlesssteel razor-blades are coated with p.t.f.e. (Polytetrafluoroethylene), a plastic with excellent insulation properties, and they can be used to construct a more advanced type of head, the sandwich type shown in Fig. 3, which will give a frequency response second to none.

This head consists of three blades clamped together, the inner one being insulated from the outside two. The signal is fed to the inner blade and the outer two are either earthed or connected to a separate bias supply (in which latter case no bias in series with the signal is needed). When the outer blades are connected to earth, they act as shields to prevent the charge spreading on the tape, and hence the high-frequency response is improved.
The use of the two shield electrodes to carry


Fig. 3: Electrostatic head made from razor-blades.

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the bias supply is a very recent innovation, and is almost an exact electrostatic equivalent of the "cross-field head" for magnetic tape recorders. The cross-field head, invented in the United States and used on several professional-quality tape recorders, the Japanese Akai models in particular.

The use of the shields greatly increases the capacity of the head, and a suitable driving circuit must be used. A cathode-follower is not wholly suitable; the capacity of the head is so high that the cathodefollower ceases to follow on negativegoing signals due to the valve cutting off.

This could be overcome by using a power valve and having a very high standing current. Another, better, method is to use the circuit sometimes referred to as the "super cathode - follower", shown in Fig. 4. Since one of the two valves must be driven on at any given time, irrespective of the polarity of the input signal, the output impedance remains low at all times.

The shielded head should not be used for


Fig. 4: Super cathode-follower. replay, as the high capacity causes a considerable loss of signal at high frequencies.

Some notes on the construction of the shielded head may be of interest. Some selection of razorblades may be necessary to find three which will not short to one another when glued together. If the blades available have a poor coating, a very thin mica shim may be used as a spacer at the blade end in addition to those used further up.

Mica sheet is very readily split into very thin portions, and with some practice, shims of 0.0001 in . can be produced. The best technique of producing such shims is to use a sharp needle to split a piece of good-quality mica at one edge. A drop of water should then be run down the needle into the split.

The water will spread between the natural layers of the mica and assist in the splitting operation. The use of water in this way also helps to prevent trouble caused by the needle crossing between layers. The mica sheets should be gently slid apart when the needle has been passed between them all over the area of the sheet.

The mica shims should be well dried before use, as they tend to retain water. The mica used must be clean and fresh; mica from an old electric iron element is useless, as it is brittle and cannot be worked readily.

The contacts to the blade should be soldered on before assembly. Stainless steel is difficult to solder, and a very hot iron is essential. The outer blades are connected together, and the inner is kept separate; remember to check the insulation between outer and inner after assembly.

The glue used should be good quality polystyrene
cement, although "Araldite" is more suitable if the blades can be kept in a suitable clamp while the adhesive sets.

Any normal tape drive from a magnetic recorder is suitable but the tape must press only very lightly against the blades, for obvious reasons. Uncoated tape is available from any manufacturer of magnetic tape (to special order) or from British Visqueen Ltd. (Acetate tape) or Dupont (Mylar tape).
" Some recent work has indicated that the permanence of the recordings can be improved by neutralising the excess charge on the tape by passing it through a "bath" of positive ions. This is done by creating a corona discharge near the tape after it has passed the recording head, although other methods such as a radioactive source can be used.

A needle held in a block of rubber and connected by suitable e.h.t. cable to a power supply (such as the e.h.t. generator of a TV set; an old set can be bought for far less than the price of building an e.h.t. supply) of at least 10 kV will give a sufficiently brisk corona for this purpose.

## GRID DIP OSCILLATOR

-continued from page 181
the medium of transfers. Finally, the pencilled calibration marks may be erased and the perspex fixed with PK screws.

## Uses of the G.D.O.

The uses of these devices are already well known but, briefly, the unit may be used for setting up the tuned stages in either transmitters or receivers, etc., without even having to switch them on! The g.d.o. is merely brought close to the circuit being checked and carefully tuned until a sharp current dip is noted on the meter.
If no dip occurs, the coil in use is the incorrect one or inadequate coupling is taking place. Immediately a dip is noted, the g.d.o. is withdrawn and carefully retuned until only the merest detection of dip is possible. The scale is then read.

As a signal generator the unit may be placed close to the aerial lead of a receiver and if a modulated signal is required, the output from an audio generator may be injected at J 1 and will be heard when the receiver is suitably tuned. Harmonics of the signal generated by the g.d.o. will also be tuneable and can also be made use of if required.

The g.d.o. may also prove useful for making c.w. or s.s.b. transmissions intelligible on a receiver not fitted with a b.f.o. No physical connection between receiver and g.d.o. is necessary to do this. Nor is modulation required: the g.d.o. is tuned close to the frequency of the signal sought and front-end injection results.

The g.d.o. may also be used to check the resonance points of aerials. Removing the h.t. supply to the g.d.o. as mentioned earlier enables the device to be used as a 'phone monitor or as an absorption wavemeter or r.f. indicator. In these cases radiated radio frequency is detected.
In conclusion it can be fairly stated that this g.d.o. is well worth the trouble entailed in its construction; it will, quite definitely prove an attractive addition to many stations.


#### Abstract

Albania: Radio Tirana (Rue Ismail Quemal, Tirana) has been reported with English at $0000-0030$ on 7,265; $0230-0300$ 9,520; 0630-0700, 2000, 2200-2230 $7,265 / 9,390$. One report says the 2000 TX is on 7,150 .

Algeria: Radiodiffusion-Television Algerienne (21 Boulevard des Martyrs, Algiers) has English from $2200-2230$ over $890 / 1,304 / 6,175$. Arabic is now being carried in the afternoons over 9,510 .

Brazil: Radio Bandeirantes (Casillon Postale 372, Sao Paulo) is reported drifting around 11,917 .


Colombia: Voz Bogota (Aereo 13018, Bogota) has been heard at 0045 on $\operatorname{HJCF}(5,960)$.

Clandestine: Radio Espana Independente can be heard between 1600-1700 on 17,695 in Spanish. Some reports say this station is located in Rumania. Has anyone any further details?

Congo: Radiodiffusion Ufac (Boite Postale 97, Elisabethville) has moved to a new frequency of 5,033 .

Czechoslovakia has, according to the International Short Wave Club, stopped jamming. Countries still engaged in jamming and to whom the club's antijamming campaign applies are Bulgaria, China, German Democratic Republic, and the U.S.S.R. occasional jamming by Hungary, Portugal and Spain.

Holland: Nederlansche Radio Unie (P.O. Box 150, Hilversum) is reported to have started a new home service transmission, Hilversum III on 1,250 .

Radio Nederland Wereldomroep (P.O. Box 222, Hilversum) has produced an English-Spanish DX vocabulary to assist Dx'ers listening and reporting to South American stations. It is obtainable free on writing, as is the printed material for the latest Dutch by Radio course. Lessons are broadcast during English transmissions on Wednesdays. The English beam to West Africa from the Bonaire relay in the $19 \mathrm{~m} . \mathrm{b}$. is now at $2130-2220$. The European relay of the 1430-1520 and 1900-1950 English transmissions is now on 6,020. The 2000-2050 English transmission is now aired in the 25,31 and $49 \mathrm{~m} . \mathrm{b}$. and the $2100-$ 2150 English transmission is in the 19 and 25 m.b.

Monaco: Trans World Radio (Rue de la Poste 5, P.O. Box 141, Monte Carlo) is now using 5,955 for its 11145-1215 French transmission. There is bad interference with Radio Liberty which also uses the same channel.

Peru: Radio Cuzco (Montero 114, Cuzco) has been heard at 0015 on the new frequency of 6,250 .

Poland: Radio Warsaw (Warsaw) has made frequency changes in the following English transmissions 1930$20001,502 / 5,995 / 6,135 / 7,125 ; 2230-23001,502 / 5,995 /$ 7,270; 2303-2330 818. The multilingual concert programmes at 1500-1630 and 2330-0100 are now on 1,502/5,995 and 1,502/7,125/7,270 respectively.
Portugal: Radio Portugal (Rua do Quelhas 2, Lisbon) has made frequency changes in the following transmissions: $0730-090021,495 / 17,740$ or 17,880 or 17,890 or 17,895; 2015-2100 6,025/7,285; 0300-0345 5,985; 0400-0445 6,025/6,185.

[^4]Uruguay: Radio Sarandi (Corporacion de Publicidad SA, Enriqueta Compte y Rique 1282, Montevideo) has been heard over CXA68 11,885 around 2215. Identifies "Noticia Radio Sarandi" every quarter hour.
U.S.S.R.: Radio Vilnius (Lietuvas TSR Radijas, ul Kanarskio 49, Vilnius) broadcasts in English on Mondays and Fridays at 2100 on $665 / 1,106 / 1,554$ and 2230 on $665 / 1,106 / 1,554 / 5,900 / 7,200 / 7,400$. The North American English service of Radio Moscow is relayed daily from 2300-2330 over 7,185/7,300.

Venezuela: Radio Juventud (Apartment 567y 576, Barquisemeto) can be heard around 2230 over YVNK, 4,900. Ondas Populares (Apartmentado 2057, Caracas) can be heard at 0035 over YVKF, 4,880. Radio Cultura (Apartmentado 1931, Caracas) can be heard around 0030 on YVKD 5,050. Radio Nacional (P.O. Box 3979, Caracas) can be heard on YVSC, 9,640 at 2400.

Reporters this month were D. Kennedy, D. A. Lavender, Middlesbrough High School S.W. Club, B. Burling, D. Mines, G. Roberts, and G. Lamb.

The Avo Multiminor Mk4 is the latest version of this well-proven multi-range measuring instrument. Designed and assembled to high standards of reliability, the Avo Multiminor offers sïmple yet instant range selection with a single rotary switch. There is only one pailr of sockets for all measurements, and the scale plate is clearly marked for easy reading.

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Also operate with two speakers on Radio. Chassis size: $15 \times 7 \times 6 \frac{1}{2}$ in high. Dial Also operates with two speakers on Radio. Chassis size: $15 \times 7 \times 6 \frac{1}{2} \mathrm{im}$ high. Dial
cream and red. $15 \times 3 \mathrm{~m}$. Valves: ECC85, ECH81, EF89, $2 \times \mathrm{ECL} 86$, EM84 and
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APATCHY period for the Amateur Bands this month, with conditions varying quite a bit at times. All the top band sleuths appear to have hung up their trusty headphones in spite of the DX still popping up on this band. Even a simple t.r.f. raked in GM, GW, and quite a few Europeans and these with only 20 to 30 ft . of 32 s.w.g. enam. wire for an aerial.

At the other end of the spectrum the 10 metre band has stirred a bit more and was open for some eight consecutive days. In a couple of years I prophesy this band will hold more DX than 20, and most probably on phone too. Just ask anybody who remembers the last sunspot max. on ten.

Fourteen and 21 megs. still provide most of the more exotic stuff. Twenty metres at the time of writing, wakes around 0800 and still has activity at midnight. Fifteen metres usually peaks between 1400 and 2000 , though it's always worth a listen just in case.

Eighty and 40 don't have such a good following. Most people who listened reported G's and EU's with some $W$ activity between 3.8 and $4.0 \mathrm{Mc} / \mathrm{s}$. The DX is comparatively easy on 20 and 15 but just try the l.f. bands and see how you really rate as an SWL.

## Low Frequencies

No reports for $1.8 \mathrm{Mc} / \mathrm{s}$ this month (one minute's silence please with heads reverently bowed). Eighty not much better. Francis Breame (Liphook) 19 set, 50 ft . l.w. reports numerous G's, DJ, DL etc., J. Hutchison (Blackpool), CR100, PR30, a.t.u. 40ft. 1.w. also reports most of Europe including DJ, DL, EA4, ON, OZ, VO1DN, VE1AOL. On 7Mc/s L. Jackson (Manchester) R1155B, 120ft. 1.w. logged LZ2KLC, KP4TIN, K3MTK, K3UKZ, UF6LA, VE1OU, VP7NQ, W1ZW, W2LXK, WA4NXC, all c.w. between 0025 and 0235 . E. Goonan (Manchester) 19 set, 50 ft . end fed, heard most of Europe on $7 \mathrm{Mc} / \mathrm{s}$, including a W1 calling "CQ 10." (Yes, I wonder, too!) the best for the session was CR4AB on c.w.

## Fourteen and Twenty-One

All sorts of gear pulling in all sorts of stations, those two bands are a hive of activity. If you only hear Europe on these bands-take up knitting! And as I cast off the last row of a jumper, let's see what the "sharp of ear" have been up to. Chis Claydon (Fife) 840C, 60ft. end fed, 20 metresCE3UT, CO6PH, CP5AQ, EA8EY, HC2SB, HK7UL, HP1BR, K7UW, KZ5LC, LU801, OA4NVE, TI2PZ, VE5US, VE6AAA, VK3AHQ, VK5TG, VP9FX, 6Y5AR. 15 metres-EA9AD, FL8MC, HM1DR, JA- 1LPZ, 2HO, 3EGE, 4BJO, 6TL, 9AMJ, KICAU/KG6 (Guam) PY7AC/ $\emptyset$, UAØLL, VK2EO, VK6RU, VP7NN, VS6FK (Hong Kong), VU2FN (India), VU2GC, YN6BF (Nicaragua ZS3XG, 606BW, 6W8DD, 9L1HL, 9M2BM, Dave Skidmore (Belper) HE4O, 20 metre dipole, 20 Metres-CN8MD 58, CR6CN 56, EA6AR 59, EP2AX 58, HR1SO 59, HS1AK/P 58, IS1VAZ 59, KP4AST 58, KR6UL 57, KX6BQ 56, LA3JM/P 59, MP4BFU, OHØNJ 59, OX3LP

59, OD5EE 58, OY7S 44, TF3UA 57, UA9EU 58, VE1AED/SU 57, VE1ADL 58, VP7NA 58, VP9BN 58, XW8AZ 58, YN1RA 58, ZB2AO 59, ZD8J 57, ZS1TZ 58, W5HWR/VP9 58, 4X4FQ 58, 5AITS 59, 5A3TB 59, 5A5TJ 59. On 15 metres Chris heard CN8FF 59, CX5AAN 57, HK8DQ 58, IT1GAI 59, KH6FBG/3 58, KP4MXN/MM 58, KV4CX 59, LU8AEF 57, PY1PAD 58, PY2AHM 58, SV1BH 59, YV5BPJ 58, 7X2AH 59, 9Q5DA 57. The two numbers after each call are the standard RST code to give an indication of how these signals are arriving at the earphones. In the $\log$ for 15 which follows the G calls logged are DX! The receiver is a 1950 Pye radiogram, the aerial is 50 ft . end fed, and the QTH is South AfricaCR6HH, CR7IZ, DJ8WP, G30AW, G3SMH, G8WPP, I1BVZ, IT1JR,' K2LBB, MP4BBA, W8HRV, W9MOD, VK6QL, ZE1BP, 5N2FEL, 5R8AL, 7Q7LC, 9Q5WO. Tnx P. Elliott for the report. Steve Wilson (Ossett) CR45 t.r.f., 130 ft . l.w., reckons twenty is bursting, he reports good sigs from--BV1USA, CR6UL, CR7IZ, CR9AM, EA8AM, EP2AX, ET3USA, HS1AK, JA1SBF/MM, JA5CC, JA6BEE, KR6QW, LA21K/P (Jan Meyen Is), MP4BCC, OA4RQ, OD5LX, PJ2ME, 'PY's, PZ1K, SVØWJ (Crete), SV1CC, T12MY, UAØKAE, VR4CN, 4X4QI, 5A1TZ, $9 \mathrm{~K} 2 \mathrm{AM}, 9 \mathrm{~J} 1 \mathrm{AB}, 9 \mathrm{M} 4 \mathrm{LP}$. Not bad for a t.r.f.? Wait till you see the 15 metre $\log -\mathrm{CM} 1 \mathrm{AR}, \mathrm{CN} 8 \mathrm{MI}, \mathrm{CR} 4 \mathrm{BB}, \mathrm{CR} 6 \mathrm{FE}, \mathrm{CX} 9 \mathrm{AAN}$, ET3USA, FS7RT, EA8ER, G3BID/CN/M, HK2AG, KP4BFF, KV4CK, KZ5SN, MP4BBA, OD5EL, PZ1BE, PY2AIR, FG7XX, SV1AB, SVØWJ, WA4PXP, WA5KKM, WN2TIB, YV7AJ, ZC4GB, ZE8JV, ZS1FT, ZS5AK, ZS6MM, 4X4QW, 5A1TZ, 5A3CAA, 5N2AAF, 5R8CR, $6 \mathrm{~W} 8 \mathrm{DD}, 606 \mathrm{BW}$ (Somalia Rep.), 9H1AD, 9J2IE, 9G1FL. Anthony Watts (Tenbury Wells), 9J21E, 9G1FL.

## Ten Metres

Paul Baker again, reports CF7FR, F9DL, many G's, LU2ADP, W's, ZC4TX, ZS1BV. Chis Claydon, too heard ZC4TX, ZS6AAC, ZS6DF, 7Q7RM, 9Q5LG, 9V1LP. C. Clarke (Farnham), 12 valve hombrew plus panoramic adaptor, folded ground plane, logged CE3PT, CN8MI, CR4BC, CR6AN, CR7IZ, ZE-2JA, 2JE, 3JO, 8JJ, ZP5KT, ZS1BV, JA, JH, 2OM, 4OI, 4PU, 6AYI, 6DK, 5A3TX, 5H3JJ, 5X5JK, 7Q7RM, 9J2-DT, RO, VX, WR, 9Q5-HD, JW.

## Next Month

VR6TC (Pitcairn) 15 metres is around but takes a bit of catching. VK9GN is in New Guinea with the Wycliffe Bible Translators. He runs 150 w . c.w. to a ground plane. (We know he's real too-Chris Claydon has a QSL from him). VK9PL (Papua) has been coming through in UK at 5 and 9 plus on 15 metres. Congrats to G3DYY for winning the fourth RSGB 7Mc/s DX Contest with a score of 2,342 points on c.w. Congrats for winning the phone section to GI3CDF who notched up 2,350 points. Contests for June include 4-5th National Field Day, 19th D. F. Qualifying Event, July 3. Fourth $144 \mathrm{Mc} / \mathrm{s}$ Contest (portables). Deadline for this month's logs is June 26.


0NE great advantage of the v.h.f. bands is that aerials are small yet highly directional and efficient. To make best use of these 'aerials it is desirable that they be made fully rotatable so that they can be turned to increase the strength of the station being received or to eliminate unwanted signals.

However, if the aerials are placed on an outside mast they become difficult to rotate as a fairly powerful motor is required and the mast must be strong and well guyed, all of which adds considerably to the expense.

In order to keep down the cost while retaining the facility of rotation, the author decided to place the aerial in the loft. Although the signal strength is reduced, no new mast is required and a simple rotator suffices.

The first attempts at a system for use in the loft were not very successful. Several motors were tried, which either failed to turn the beam or turned it too fast. Power supplies also had to be found for the low voltage d.c. types. As no indication of the beam's direction could be simply obtained, the idea of a motor-driven system was abandoned in favour of the system described.

No motor is employed, the beam being turned by a single continuous belt which passes through two $\frac{1}{8}$ in. dia. holes in the ceiling. The advantages of this system, beside minimal cost, are variable speed so that the beam may be quickly turned and yet accurately set, and simplicity in indicating the beam's direction.

The beam is supported on a 1 in . dia. wooden pole sold as a broomstick, between two convenient joists. The bearings at both ends consist of pieces of $\frac{1}{4} \mathrm{in}$. dia. brass or steel rod. These are drilled and countersunk (see Fig. 1), and set in the joists.

The other part is formed by $\frac{1}{4} \mathrm{in}$. rod ground
to a point. This is set firmly in the broomstick by drilling $\frac{1}{4} i n$. holes in the ends and making two cuts at right angles to enable the wood to be squeezed on to the rod with a Jubilee clip (see Fig. 2).

A bicycle fixed-wheel sprocket is fitted to the $\frac{1}{4} \mathrm{in}$. rod and this is driven by a chain attached to the belt. This is the only difficult part of the job, the part which calls for ingenuity on the part of the individual constructor, as the sprocket must be fixed concentrically to the $\frac{1}{4} \mathrm{in}$. spindle. The problem is that the size of the hole in the sprocket is much larger being $1 \frac{3}{8} \mathrm{in}$. diameter.


Close-up of author's two-metre rotating aerial.

The best solution is to have a suitable adaptor made, but there are many small engineering firms who will undertake such work. However, the author was able to improvise a connector using an old loudspeaker magnet.

This was just the right size to fit the flange on the sprocket and the polepiece drilled to accept a $\frac{1}{4}$ in. shaft. The shaft was threaded and two nuts used to clamp the magnet. This is shown in Fig. 3, and Fig. 4 shows a suitable connector to have made up.

No doubt other possibilities will occur to readers but the solution offered, although crude, works perfectly. A different type of cog could be used provided it can be positively driven and this would eliminate the problem.

Two small pulleys are supported above the holes in the ceiling by a board between the joists, for the belt to run over (see Fig. 5).

In the shack, the belt runs around a large drum of the type used as a slow-motion drive cord drum. A large one is required to enable the beam to be swung through 360 degrees. The author's is $4 \frac{1}{2} \mathrm{in}$. dia. This is fixed to the wall by a suitable bracket (see Fig. 6). The scale indicating direction is fixed to the drum.

The belt is best made of nylon cord of the type used for curtains, as this has great strength and will not stretch. The overall length of the belt should be kept as short as possible. If a long length is required in the loft it will be necessary to use extra pulleys to prevent the chain slipping off the cog.

A large spring is incorporated to keep the belt tight. This is best placed between the chain and belt as it then provides a convenient way of shortening the belt by hooking it on another link of the chain.

The author has used this arrangement to turn a four-element beam for two metres. This it does easily and could obviously turn a larger array if required, the maximum size being governed most of all by the space available in the loft.


Part of pulley which projects beneath the ceiling.


Fig. 1: The fixing ends for the centre rod which carries the full weight of the aerial.

Fig. 2: A method of securing the brass or steel rods to the ends of the centre rod.

Fig. 3: Author's method of fitting the sprocket to the metal rod at the lower end of the centre rod.

Fig. 4: Enlarged view of the clamping plates for the sprocket.
Fig. 5: Two pulleys attached to a sheet of hardboard for guiding the Nylon belts through the ceiling.

Fig. 6: Lower pulley, showing the fixing arrangements.


WITH its five ranges, this transistorised oscillator tunes from $5 \mathrm{c} / \mathrm{s}$ to $75 \mathrm{kc} / \mathrm{s}$ by variation of the resistances and capacitances of a Wien network. This form of tuning has insufficient selectivity for the suppression of distortion unless the amplitude is closely controlled, and this function is taken over by a sensitive thermistor which maintains a constant amplitude and the maximum output is at a level of 1 volt r.m.s. independent of frequency.

## Scales

Linear potentiometers if incorporated for the fine tuning produce a crowding of readings at the high frequency end of the scales and the tuning becomes correspondingly more critical in setting. This is satisfactorily overcome through obtaining ganged potentiometers wound to an inverse semi$\log$ characteristic, and the result is an approach to an ideal frequency scale on which octaves are represented by nearly equal distances along the scale.

A simple form of transparent cursor is employed on the tuning dial, its straight edge serving as a

## YYIDE REA

## BY R. LEYLAND

ruler for initially marking in the frequency divisions, and subsequently as an indicating-line sweeping the five scales, but a hair-line cursor could easily be substituted if preferred. Two limit positions are first marked on the dial, which is temporarily secured in place with a few spots of adhesive, and act as reference points for aligning the cursor when the knob is being refitted.
The cursor is fastened on the knob by a 1 in . aluminium disc countersunk for the three 8BA screws, and there is a thin washer underneath, on the $\frac{1}{4}$ in. shaft, preventing contact with the dial. The small clearance between the cursor and dial avoids scratches and does not introduce any appreciable parallax in readings.

After calibration in pencil, the dial is removed for inking-in, and is covered with a clear plastic material. Then it is replaced and cemented permanently in position. Small countersunk screws could be used instead, but if any of the paint situated beneath the dial is affected by the adhesive, it can be scraped off and more adhesive successfully applied.
The dial and escutcheons are of aluminium to which drawing paper is bonded with adhesive; the circles, etc., for the scales having first been inscribed with Indian ink. On completion of the scales, some preliminary experiment is advisable before proceeding to cover the dial with the clear plastic. A clear adhesive is spread thinly with a circular motion and allowed to become almost dry before pressing the plastic material down on top.

## Circuit Description

The oscillator circuit uses p-n-p transistors throughout and consists of a feedback loop around transistors $\operatorname{Tr} 1$ to $\operatorname{Tr} 4$. The fifth transistor $\operatorname{Tr} 5$,


Fig. I: Theoretical diagram of the audio oscillator.

## GE A.F. OSCILLATOR

an extra stage to feed the step attenuator, increases the total battery current to 25 mA at 9 volts. The circuit works on battery voltages up to 12 volts, at which the current is about 35 mA . When the voltage falls as low as 6 volts there is a tendency towards distortion.

Oscillation is at the frequency of zero phase shift in the Wien network, which delivers one-third of its input votage to the base of Tr1. However, the actual input to $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$ is reduced by feedback via thermistor type R53. The thermistor keeps the output across VR2 close to 1 volt r.m.s. by controlling the amount of negative feedback. Any excess of output above 1 volt r.m.s. is transferred by the thermistor to R7 where it becomes negative feedback to reduce the input to the amplifier.

Although power-actuated, the thermistor is effectively a voltage-controlled resistance with a gradual response to changes in the r.m.s. value, and it takes over the task of keeping the output at a constant level. Its response, too slow to cause distortion


FIg. 2: Oscillator chassis (lower side)
even at the lowest frequencies, cannot cope immediately with transients such as that caused by switching on, and it therefore takes a moment or two to settle dewn.

Direct current is kept out of the thermistor by feeding it from the output side of C12. The emitter of $\operatorname{Tr} 4$ is also a low-impedance driving point for the Wien network and is able to preserve a constant a.f. voltage across it, despite large changes of network impedance with tuning as the ganged potentiometers are varied between their minimum and maximum value. High values of capacitance for C12 and C14 provide a low coupling impedance down to subsonic frequency, but except for C11 the positive feedback loop is direct-coupled, and C11 feeding the high input impedance of $\operatorname{Tr} 3$ need not be so high in value.

To avoid loading the Wien network, the amplifier has a very high input impedance, produced by the emitter-follower, Tr 1 , which feeds Tr 2 , which also has its input impedance raised by the emitter resistor R6. The bias resistors R2 and R3 are in


Fig. 3: Oscillator chassis (upper side).
effect part of VR1b and do not shunt the input impedance of the transistor stages. An adequate current in $\operatorname{Tr} 1$ is ensured by an emitter resistor R4. This cannot be omitted altogether, but in the more usual arrangements R4 would have a much higher value and would be connected to the +ve line instead of to the emitter of Tr2. It works quite well as shown however, with high gain transistors.

The amplifying stages $\operatorname{Tr} 2$ and $\operatorname{Tr} 3$ have negative feedback applied locally by the emitter resistors. There is also the overall negative feedback via the thermistor, which is cancelled out at one frequency by positive feedback from the Wien network. The amplification is adequate to maintain the stabilised output up to about $80 \mathrm{kc} / \mathrm{s}$ on range 5 at which point oscillation ceases. A small capacitor across

R11 would maintain the oscillation right up to the end of the scale on range 5, but this shifts the calibration, and it seems undesirable to impair the performance in order to extend the coverage above $80 \mathrm{kc} / \mathrm{s}$ in an oscillator that is intended primarily for audio frequencies.
Tr4, an emitter-follower output stage, must have the full signal-handling capacity, so it is biased to midway between zero and the -9 V supply line, and as it has a low emitter load resistance, it will take a current of some 10 mA . A further output stage, Tr5, is included to feed the switched attenuator without appreciable loading on VR2.

## Attenuator

Accurate attenuation is less important than that it should be independent of frequency, because in measurements on the gain of amplifiers, it is probably better to rely upon an a.f. voltmeter of the thermionic or transistor type, with its own multirange facilities.
The continuously variable attenuaton of VR2 can be calibrated at $50 \mathrm{c} / \mathrm{s}$ using a rectifier voltmeter of fairly high impedance, and does not appear to be quite linear. The rectifier voltmeter was carefully checked against an accurate moving iron type supplied from a transformer, using low resistance potential dividers to obtain smaller voltages.

For larger attenuations there are fixed steps. Ideally the step attenuator should possess a constant output impedance, so that the insertion loss on connecting a load remains the same at every step. The impedance should be unaffected by the output transistor, which must therefore offer low output impedance. However, it is equally possible to utilise a very high output impedance, which is available at the collector.

Instead of a constant impedance, the attenuator shown in the oscillator circuit, Fig. 1, has a low impedance, about $80 \Omega$ at the 100 mV step and less at the other steps. At the 1 V position of the attenuator, loads of resistance less than $1 \mathrm{k} \Omega$ should not be connected if distortion is to be minimised,
but at the other attenuator steps, any value of load can be connected.

The low resistance attenuator elements had to be constructed by winding them on cards. These are of 0.024 in s.r.b.p. sheet, which can be cut with scissors, and the edges are smoothed with emery cloth. The resistances, of the non-inductive type, are made by winding two wires simultaneously in opposite directions on the card, exchanging the bobbins between hands at each half turn. The card is mounted vertically, and connections are made as in Fig. 8. One resistance wire is adjusted to twice the resistance value. Then the other is connected and adjusted to give the correct value. One of the resistors has the value $1 \cdot 11 \Omega$, which in parallel with $10 \Omega$ yields a value of $1 \Omega$.

To protect them against damage, a box of insulating material is fitted over the resistances, and fixed to pillars with 6BA screws. The entire attenuator is constructed as a separate unit from the oscillator chassis, although mounted beside it. Data for the wire-wound resistances is as follows:

| Resistance | Wire <br> (Constantan) | Lengths | Card Size |
| :---: | :---: | :---: | :---: |
| $1 \Omega$ | 30 s.w.g. d.s.c. | Two 36 cm. | $1 \times \frac{5}{8} \times 0.024 \mathrm{in}$. |
| $1 \cdot 1 \Omega$ | 30 s.w.g. d.s.c. | Two 38 cm. | $1 \times \frac{1}{16} \times 0.024 \mathrm{in}$. |
| $9 \Omega$ | 38 s.w.g. d.s.c. | Two 74 cm. | $1 \times \frac{11}{16} \times 0.024 \mathrm{in}$. |
| $90 \Omega$ | 44 s.w.g. d.s.c. | Two 2 m. | $1 \times \frac{7}{8} \times 0.024 \mathrm{in}$. |
| $900 \Omega$ | 47 s.w.g. d.s.c. | Two 8.5 m. | $1 \frac{1}{4} \times 1 \frac{3}{8} \times 0.024 \mathrm{in}$. |

An appreciable amount of work is involved in the construction and adjustment of these resistances, and a very adequate alternative form of attenuator can be made as in Fig. 7, using high stability resistors of $5 \%$ tolerance. These are made in preferred values, and the values selected give the attenuator a constant output impedance of approximately $95 \Omega$ with the same attenuation ratios as before-i.e., 10:1. Altering the voltage across a load impedance by a factor of 10 alters the power by a factor of 100 (or 20 db ), so these can also be described as 20 db steps.

## Construction

The oscillator chassis consists of a piece of $7 \frac{1}{4} \mathrm{x}$ $3 \frac{3}{4} \times 1 / 16 \mathrm{in}$. paxolin, suitably drilled to take push-fit eyelets, and the case is made from 18 s.w.g. aluminium.
The larger capacitors are secured on the chassis by soldering their leads into eyelets. Smaller capacitors, mounted around the miniature wavechange switch S1 are connected to two junction points formed by soldering short pieces of wire vertically into eyelets. (See Fig. 2). Interconnections on the reverse side of the oscillator chassis only require sleeving where they cross as shown in Fig. 3.
The oscillator chassis is mounted on $\frac{5}{8} \mathrm{in}$. spacers with 6BA countersunk bolts, $\frac{7}{8}$ in. long, through the top of the metal case. Shakeproof washers are placed under the 6BA nuts on the chassis. The attenuator is mounted in the same way.
components list

| Resistors: |  |  |
| :---: | :---: | :---: |
| RI $2 \cdot 2 \mathrm{k} \Omega$ | R8 | $56 \mathrm{k} \Omega$ |
| R2 $15 \mathrm{k} \Omega$ | R9 | $10 \mathrm{k} \Omega$ |
| R3 $2 \cdot 7 \mathrm{k} \Omega$ | R10 | $3.9 \mathrm{k} \Omega$ |
| R4 Ik $\Omega$ | R1I | $1 \mathrm{k} \Omega$ |
| R5 $2 \cdot 2 \mathrm{k} \Omega$ | R12 | $470 \Omega$ |
| R6 $470 \Omega$ | R13 | $5.6 \mathrm{k} \Omega$ |
| R7 $100 \Omega$ | R14 | $10 \mathrm{k} \Omega$ |
| Capacitors: |  |  |
| $\mathrm{Cl} \quad 1 \mu \mathrm{~F} 125$ |  |  |
| C2 $0 \cdot 1 \mu \mathrm{~F}$ |  |  |
| C3 $0.01 \mu \mathrm{~F} \mathrm{125V}$ polyester |  |  |
| C4 1000 pF tubular ceramic |  |  |
| C5 100 pF tubular ceramic |  |  |
| C6 $1 \mu \mathrm{~F} 125 \mathrm{~V}$ polyester |  |  |
| C7 $0.1 \mu \mathrm{~F} 125 \mathrm{~V}$ polyester |  |  |
| C8 $0.01 \mu \mathrm{~F} 125 \mathrm{~V}$ polyester |  |  |
| C9 1000 pF tubular ceramic |  |  |
| C10 100pF tubular ceramic |  |  |
| CII $100 \mu \mathrm{~F}$ I5V electrolytic |  |  |
| CI2 $1000 \mu \mathrm{~F} 15 \mathrm{~V}$ electrolytic |  |  |
| CI3 $100 \mu \mathrm{~F}$ I5V electrolytic |  |  |
|  |  |  |
| Transistors: |  |  |
|  |  |  |
| TrimTr5 OC45 Mullard |  |  |
| Miscellaneous |  |  |
| Thermistor type R53 (S.T.C.) |  |  |
| VRI $25 \mathrm{k} \Omega$ dual-ganged potentiometer, both sections inverse semi-log. Reliance type TW. |  |  |
| VR2 $\mathrm{k} \Omega$ non-inductive potentiometer, Reliance type TW. |  |  |
| SI, S3 2 pole, 2 way rotary switch. |  |  |
| S2 On/off | h. Fl | coaxial socket. |



Fig. 6: The tuning knob and cursor.

Fig. 7: Wirewound attenuator resistance.


Fig. 8: Alternative form of attenuator.
Frequency calibration of the oscillator by using an oscilloscope is greatly facilitated by temporarily setting up an auxiliary oscillator. Next month details will be given of a special form of auxiliary oscillator which produces an elliptical trace.

## Ex Service Equipment

No doubt many readers will have experienced difficulty and sometimes frustration in trying to undo nuts and screws fitted to this equipment. These parts are usually varnished around the nuts and screw ends and it is almost impossible to remove them in the normal manner without breakage.
It has been suggested that a hot soldering iron applied to the nut will do the trick, but this method is clumsy and not entirely satisfactory.
The best way I have found is to dissolve the varnish by applying a little cellulose thinners, and after a few minutes, the nuts can be easily removed with a suitable box spanner. Moreover, they can be used again and there is no risk of damage to any components that one wishes to salvage.
F. B. Poppitt.

> Bearsden, Dunbartonshire.

## Single Circuit Panels

Mr. R. G. Young's letter in the June issue of Practical Wireless referred to my article "Single Circuit Panels" and asks why I assume that copper cladding has to be used.
I do not. The idea of using wire for panels of similar purpose was fully described in my article "Bread-board Wiring" published in the January 1965 issue.

## W. Groome.

Halesowen, Worcestershire.

## Add-on B.F.O.

I have just fitted an "Add-on BFO" as described in your October 1965 issue to my transistor set (Lasky's Radio Skyrover with extra audio stage). I found that in order for the circuit to operate correctly, it was necessary to wire a $0.01 \mu \mathrm{~F}$ capacitor across the $10 \mathrm{k} \Omega$, R1 resistor.
M. J. Draycott.

## Hitchin, <br> Hertfordshire.

## Anyone a Genius

With reference to the letter from Mr. Wright in last month's Practical Wireless I have an idea which may perhaps interest this gentleman and other readers.

I also became interested in the "fuzz box", but did not really take to the idea of distortion. I found that instead of this a relatively high frequency tremolo unit gave a similar effect.
The transistor tremolo is simply a preamplifier with its source of power fed from a multivibrator circuit at $5-15 \mathrm{c} / \mathrm{s}$. This uses discharge capacitors of about $2 \mu \mathrm{~F}$. If these were decreased to say $0.5 \mu \mathrm{~F}-1 \mu \mathrm{~F}$ and the rest of the circuit correspondingly a suitable frequency would be obtained. M. Gilbert.

Crayke,<br>York.

# NEWS AND.. 

## U.S. SHOWGASE IN BRITAIN

The American stand at the Instruments Electronics and Automation Show, just ended at Olympia, London, represented the biggest ever export promotion undertaken by the United States' electronics industry in Britain. The stand was also the largest single display at the exhibition, showing products from some 72 of America's leading manufacturers in the field.

America's space programme has given her electronics and computer industries fresh impetus during recent years and many of the exhibits illustrated the commercial and industrial applications which have resulted from equipment and techniques developed during space research.

## GERMAN HI-FI EOUIPMENT



The turntable shown above, complete with stereo pick-up, is made by the German firm of Braun A. G. and costs $£ 187$. It is part of Braun's new Studio 1000 range of hi-fi equipment, which also includes tuner unit ( $£ 338$ ), stereo control amplifier ( $£ 345$ ) and speaker system ( $£ 523$ ).

All these units have been designed and manufactured to the highest standards to achieve faithful sound reproduction. The tuner (CE 1000) and the amplifier (CSV 1000) are both transistorised. The tuner covers I.w., m.w., s.w. and f.m. and incorporates switchable a.f.c. which is automatically turned off when the hand touches the separate f.m. tuning knob. There is also a built-in tuning meter.

The control amplifier, which includes a separate preamplifier within the same cabinet, provides 55 W per channel continuous output with distortion below $0.3 \%$ at $\mathrm{Ikc} / \mathrm{s}$. There are more than the normal number of controls and the five inputs feature variable sensitivity.
The L 1000 speaker system was designed to meet standards set by the German Institute for Radio Technology. In one enclosure Braun house three woofers, eight mid-range speakers and two tweeters.

The turntable (PS 1000) operates on four speeds although a fine-speed control permits exact settings, helped by a built-in stroboscope. The pick-up arm is balanced by two adjustable counterweights and turntable shut-off is achieved by a photo-electric device.

Fi-Cord International are the U.K. agents for all Braun equipment which, by the way, includes many more moderately priced hi-fi units.

## RADIO SOCIETIES AMALGAMATE

On June Ist, the South London Mobile Club, Wimbledon and District Radio Society and the Purley and District Radio Club merged to form a new association. Hon. Secretary of the new Southern Amateur Radio Association is B. Negri, G3LXN.

# COMMENT 

## PHILIPS' SUPER-PORTABLE

Philips recently unveiled some 30 new products-radios, radiograms, television receivers and record players-at a trade fair in Brighton, Sussex. As with each of these categories, radios on show ranged from the economy class to high quality; from the "Popmaster" pocket portable at $£ 7$ 19s. 6d., to the FM-AM De Luxe at...... . well, first see what you get.

It covers seven bands; l.w., m.w., four s.w. and f.m., plus an extended long wave band which permits reception of beacon transmitters, weather forecasts and time signals and trawler band coverage which picks up marine broadcasts. Adjustable a.g.c. on s.w. and a.f.c. on f.m. improves reception, as does the d.c. tuning meter. The IW output comes via a 5 in . $x$ 7in. speaker and the transistorised circuit operates from six $1 \cdot 5 \mathrm{~V}$ cells although it can be connected to a main unit. Outdoor aerials for a.m. and f.m. can be connected, but apart from the normal ferrite aerial, in-built frame and telescopic dipole antennae take care of s.w. and f.m. reception. Gramophone and tape inputs, built-in earphone, illuminated dial, world-wide time map, treble and bass tone controls and azimuth ring for navigation, sum up most of the "extras", although there are more. And the price for this $\mathbf{I} 4 \frac{1}{2} i n . \times 10 i n$. portable?-just $[00$ guineas.

## JUDGEMENT ON RADIOS

In the April issue of "Which?", the Consumers' Association journal, 25 portable radios are candidly assessed for quality, operation and value. Laboratory tests and listening panels produced a revealing set of results on radios costing from $£ 3$ 9s. 6d. to $£ 11$ Ils.

## R/Ts FOR P.C.s

"Beat" constables in six divisions of the Metropolitan Police District are now equipped with portable two-way radio telephones.
The introduction of "mobile" communications, to the Force followed some months' trial with standard equipment. The design which evolved from these tests consists of a main receiver/transmitter unit and a microphone/loudspeaker. Except when in use, the whole equipment is concealed in the policeman's uniform and the makers overcame a major obstacle by incorporating the aerial in the flexible lead joining the two units, thus obviating an inconvenient rod aerial.

The new equipment will eventually put all London's pedestrian police officers in close contact with master stations in sub-divisional headquarters and mobile $R / T$ vehicles.

STEREO TAPES FOR MOTORISTS


The new Veritone tape-player for cars and boats operates automatically when pre-recorded tape cassettes are inserted. The main unit fits beneath the dashboard of a car as shown above, and provides over 30 minutes of stereo sound from two or four speakers.

Apart from a whole library of jazz, pop and classical music, taped language lessons are also available. Cassettes cost $£ 2$ IOs. each and tape-player and two speakers, 48 guineas. Veritone Limited, are the U.K. distributors.

## S.W. Broadcast Stations

Again your magazine prints information on the stations Radio Moscow, Radio Pekin, Radio Berlin International, Radio Prague and Radio Sofia, Bulgaria. Are you and Mr. Guttridge unaware of the amount of deliberate interference caused by these stations which insist upon using $98 \%$ of all the available channels in the broadcasting bands from 16 m to 75 m .

They do not seem to be only satisfied with ordinary a.m. but also use distorted modulation and buzzing noise transmitters and they now have the nerve to operate a.m. and noise producing machines in the amateur 20 and 40 metre bands. I think that these stations should not be mentioned in Practical Wireless and when they see that they are losing support, perhaps they will do something about the interference they are causing.
N. D. Mugford.

> R.A.F. Episkopi,
> B.F.P.O. 53.

We would be interested to hear what other readers may have to say on this rather controversial point-Editor.

## Tapespondent Wanted

I should like to tapespond with any person of similar age to myself (16) who takes an interest in SWL radio in general, tape recording, special effects amplifiers, music ("pop", electronic or otherwise) or any associated subject.

I am at the moment studying for the R.A.E.
P. C. Underhill.

> Pant Mawr, Harlech,
> Merioneth, Wales.

## 4 Metres

I Read in the April 1966 issue of Practical Wireless a letter from F. G. Rayer concerning v.h.f. coils. The one he described was to tune the v.h.f. f.m. station at Wrotham on $3 \cdot 1$ metres. Quote: "but three turns about $\frac{5}{8} \mathrm{in}$. diameter, $\frac{3}{8} \mathrm{in}$. long self supporting'. I myself would be interested in application to 4 metres etc. and would therefore like to hear from other readers interested.
R. A. Adair.

13 Seaview Terrace,
Holywood,
Co. Down,
N. Ireland.

## Correspondent Wanted

I would like to correspond with anyone of my own age (15) who has built the "Versatile Gramophone Amplifier" on page 336 of the August 1964 edition of P.W. A. Kenward.

289 Longford House,
Uxbridge Road, Hampton Hill, Middx.


FIRST, let's get rid of one misconception. Automatic control of recording level is not a new technique. Many of the older dictating machines and one or two portable tape recorders employed signal-operated gain control circuits. These were, of course, intended for speech reproduction, where intelligibility is more important than absolute fidelity, and some frequency compression can be tolerated.

Lately, the leading tape recorder manufacturers have brought out models which are "automatic", but which sacrifice nothing in terms of quality. These have raised howls of protest from the tape recording enthusiasts, who regard the loss of a manual recording level control in much the same light as the motoring aficionado regrets the absence of a gear-lever in a car with automatic transmission. As a peace-offering, the manufacturers now provide an "over-riding" switch to allow either automatic or manual control.

## Quite Simple

The principle of a.g.c. is quite simple. In its essentials it consists merely of a bias applied to a controlled amplifier stage, or stages, in proportion to the level of the incoming signal. The greater the signal, the greater this bias, and the more the overall amplifying system tends to return to a steady statepredetermined by a delay applied to the bias line.

Complications ensue when we apply the bias to different stages, as in a television receiver or a quality radio; then these delays differ, and extra circuitry is needed to prevent interaction between

stages. We are not concerned with these factors at present, and reference to Fig. 1 is sufficient to illustrate the principles of a.g.c.

This is part of the circuit of a conventional radio receiver. The i.f. signal, after amplification by V1, is applied to the detector diode D1, rectified and passed, as an audio signal, to the a.f. amplifier V2.

Part of the i.f. signal is tapped off via C1, applied to the second diode, D2, rectified and applied via filter circuits as a varying d.c. bias to the grid circuit of V1. R1, the a.g.c. diode load, is returned to chassis, and the cathode of the double-diode triode valve is self-biased by R 2 , decoupled by C 2 .

But the bottom end of the a.f. detector load is returned to the cathode directly. Thus, the a.g.c. rectifier has to overcome the bias of the valve before it can begin to conduct. This provides the "delay" -which is a delay in voltage, not time. The a.g.c. does not come into operation until a certain minimum level of signal is reached.

The components, R3, C3, filter the a.f. component from the bias, and the long time constant ensures a steady bias, proportional to the average variations in signal level. The overall time constant of the circuit is an important consideration during designtaken for granted by those of us that use and repair the receivers.

For audio circuits, the time constant becomes even more important, and it is this fact that makes tape recorder a.g.c. difficult to design. Obviously, a bias voltage exactly proportional to the signal would be the same as a simple reduction in gain. (In this, and the following discussion, the term bias refers to the


Fig. I: (on the left) shows a conventional radio a.g.c. circuit and Fig. 2: illustrates the control range of the a.g.c. circuit.
a.g.c. voltage, and not the recording bias, with which we are not at the moment concerned.)

Similarly, a circuit with the same "averaging" characteristic as the radio circuit would be useless. Consider a passage of music which starts with a loud chord, or clash of cymbals. The opening high-level signal would set the bias circuits in operation and determine the level of bias, which would then die away as no further loud signals sustained it.

But the period of time over which it dies away could include several quiet sounds, which would be recorded even more quietly despite the fact that they would not normally be great enough in signal level to produce an appreciable bias. This is more clearly seen if we study a diagram of the effect of tape recorder a.g.c. on a signal, and note the time factors.

Fig. 2 shows the control range of the circuit. This is actually the curve of the a.g.c. circuit of the Philips EL3552. The vertical axis shows the sensitivity, measured in dB , and the horizontal axis represents time, in seconds.

This machine has a maximum control range of 30 dB , which is quite sufficient for normal purposes. When a loud sound occurs, the control circuit comes into operation and almost instantaneously the attenuation from 0 to -30 dB occurs.

At this point, if no further loud sounds occur, the circuit begins to recover, and the curve shows that in approximately one minute the original amplifier sensitivity is restored.

## Effects of A.G.C.

The effect of the audio signal is shown in Fig. 3, also reproduced from the Philips data on the EL3552, which was one of the pioneer models of modern a.g.c. techniques.

The hard curve shows a passage of music with a pronounced peak at B. The chain-dotted line from point $\mathbf{R}$ indicates the level ( 0 dB , see Fig. 2) at which the control circuit comes into operation. Note the attenuation period, t .

The recorded signal is now reduced from its peak at B to the level of R, and then follows the curve of the dotted line. Note next that this is virtually a replica of the hard curve. The important relationship between variations of sound intensity is maintained; in other words, distortion is not introduced.
This may seem, at first, a contrary argument to the time-constant point that was previously


Fig. 3: (above) shows the effect of a.g.c. on a musical passage. $R$ indicates the suppression level, " $t$ " the fall time, and the dotted line an uncontrolled signal. Fig. 4: (on the right) shows the form of control used on dictating machines featuring two-position sensitivity switches for record and playback levels.
discussed. To understand it more fully we need to consider the recording process. We need, in fact, to consider what we do when we set the tape recorder to its desired level by a manual gain control.

Suppose, for example, we record at a gain lower than we should. The tape will be undermodulated, and when we play this back it will be necessary to turn up the volume control to get the required output.

But this introduces noise, the inevitable problem due to electro-mechanical, and purely electronic limitations of the system. Tape hiss, input stage noise level, hum, etc., are more evident at high playback volume control settings.

## Adjusting the Record Level

If, on the other hand, we record with the gain control too far advanced, the tape is overmodulated and distortion sets in. No matter what volume control setting we then use on playback this distortion will be present. The recording level indicator, be it neon lamp, meter or magic eye, is there for the purpose of warning us when we tend to overmodulate.

In other words, it is a peak-indicating device. We set our level so that the sounds being recorded do not overmodulate the tape on peaks and know that the rest of the sound is in proportion.

This then is precisely what the automatic circuit is doing. By setting the circuit to come into operation at the level R in Fig. 3 we are ensuring that no incoming signal will overmodulate the tape, and the short fall and slow rise of the curve then gives us a recorded signal at the correct level throughout.

The setting of the level R is determined during design, and preset controls are incorporated to allow for small variations. We shall come to the difficulty of adjusting these controls as we consider the circuits.

The circuit of Fig. 4, from an early version of the popular Grundig Stenorette, is very similar to the radio circuit we have already looked at. A double-diode-pentode valve is used, with diodes strapped.

The signal is applied via a $2,000 \mathrm{pF}$ capacitor from a tapping on the secondary of the output transformer to the diode anode, and the load is the $10 \mathrm{M} \Omega$ resistor R1. Filtering is provided by R2, C2 and R3, C3, and the bias voltage is applied to the pentode section of the EBF80 and to the grid of the input stage.



Fig. 5: The left-hand circuit (a) shows the automatic control section of a transistorised dictating machine. Part of the output is fed back to bias the first transistor. The other circuit (b) shows a different method, with the emitter voltage varied by a rectified signal voltage from the output transformer.

A feature of this circuit with some bearing on the operation of the automatic control is the twoposition sensitivity switch S1. This selects the tapping position along the anode load of the first stage, from which the signal is taken to the second, and is made further adjustable by the preset resistor VR1.

## Using Feedback

A variant of this circuit is used in the later, transistorised version of the Stenorette, as seen in Fig. 5a. Here, the audio signal from the output transformer is again used, causing the diode D1 to conduct.

The difference is that control of the base bias potential of $\operatorname{Tr} 1$ has to be effected without any alteration in the load impedance or other parameters. This is done by making D1 part of the stabilising circuit of the base bias circuit, feeding the stabilising voltage from the junction of R1, R2, which are across the negative to chassis potential of the instrument.

Audio is then applied to the diode via R3, C1, and as it conducts, the base potential of $\operatorname{Tr} 1$ changes, reducing its gain.

The circuit of Fig. 5b is a refinement of this, used on the Stenomatic dictating machine, also by Grundig. Although this machine is not typical, being mechanically very different from the tape recorder types, and using coated foil wrapped around a rotating drum instead of conventional spools, its circuitry is what concerns us here, and Fig. 5 b shows it has two diodes, apparently in opposition.

Again, the prime aim is to control the bias of Tr1, but this time by altering emitter voltage. Normal d.c. stabilisation of the $\operatorname{Trl}$ base is by the potential divider R1, R2. R3 biases the diode D1 in a forward direction. When an audio signal arrives via the lowpass filter $\mathrm{C} 1, \mathrm{R} 4$, it is applied to the second diode, D2, rectified positively and used to counter the forward resistance of D1.

C2 is used to smooth out audio frequency variations of rapid periodicity and C3 is the charge capacitor. The reference level across C 3 is used to alter the bias of Trl.

Another transistorised circuit, but with more serious audio functions and therefore with a choice
of both manual and automatic recording level, is that of the Fi-Cord 202, shown in Fig. 6. When the gain control is turned fully anti-clockwise the switch S2A opens and the transistor $\operatorname{Tr} 2$ is part of the collector load of Tr1. (Note that a fixed d.c. collector load, R4, is provided, or damage to the transistor would ensue during switching.)

When the machine is in the Play condition, Tr2 is shorted out. During Record, the emitter bias of Tr 2 is set by the full resistance of VR1 (in its minimum position) and the base receives a rectified audio signal.

## Varying the Impedance

This causes it to change impedance and act as a varying a.c. load to the first stage collector, giving an automatic control of volume. In fact, with a transistor connected in this way, its impedance varies inversely with the amplitude of the signal being fed back.

The circuit is very responsive to peaks, but the built-in delay of later machines to maintain regular balance between uncontrolled and controlled levels, as shown in Figs. 2 and 3, has not been given such weight.
The extra circuitry involved in obtaining this regularity for quality work can be seen by reference


Fig. 6: In this Fi-Cord 202 circuit, a transistor is used as a varying collector load for the first amplifier: its impedance being varied by the feedback.


Fig. 7: Elements of the control circuit of the Philips EL3552. Component values are as follows; R4, $22 \mathrm{k} \Omega: R 5,68 \mathrm{k} \Omega: R 6,10 \mathrm{k} \Omega: R 7$, $22 \mathrm{k} \Omega: R 9,470 \mathrm{k} \Omega: R 10,10 \mathrm{M} \Omega: R 12,100 \mathrm{k} \Omega: \mathrm{RI} 3,100$ ohms: R14, $22 \mathrm{M} \Omega: R 15,100 \mathrm{k} \Omega: R I 7,22 \mathrm{M} \Omega: R 18,100 \mathrm{k} \Omega: R 40,22 \mathrm{M} \Omega$ : C6, $4 \cdot 7 \mathrm{kpF}:$ C7, $390 \mathrm{pF}:$ C8, 33 kpF : C10, 47 kpF : C12, 10 kpF .
to Fig. 7, which shows the portion of the Philips EL3552 circuit under discussion. It will be noted that a control triode is used as an amplifier, to obtain sufficient amplitude of signal from the output of the transistorised first stage, rather than tapping off a portion of the amplified output.

The controlled stage is an EF83, vari-mu valve whose grid circuit consists, during selection of the Auto function, of the load of the diode rectifier. The signal from the collector of $\operatorname{Tr} 1$ is fed to the EF83 grid via C10 and the manual volume control, for normal; i.e., "Manual" operation, but when the "Auto" switch is selected this part of the circuit is bypassed by Sk2 and the direct signal is via C10 and C12.

At the same time, the other section of Sk2 closes and the two series resistors R14 and R40 become part of the grid load of the vari-mu valve. The rectified output from the triode valve supplies the negative potential to control the gain of the valve and the exact choice of component values gives the delay, which is approximately one minute for microphone signals with this machine.

A similar principle is employed with the Elizabethan Automatic tape recorder (see Fig. 8). A control triode is used to amplify the signal for application to a rectifying circuit, providing a negative potential to the grid of a vari-mu valve.

The particular points to note about this circuit are the preset controls, VR2 to determine the output of the triode, and VR1 to set the recording level, and the voltage doubler circuit, D1 and D2, which gives both a larger and a better regulated bias voltage. The necessary long time constant is supplied by the combination of the
rectifier reservoir capacitor, $\mathrm{C} 6(1 \mu \mathrm{~F})$, and the load resistor, R10 ( $80 \mathrm{M} \Omega$ ).

## Adjusting the Charge Capacitor

Because of this long time constant it is necessary to use a little circumspection when testing and setting up automatic tape recorders. The charge capacitor C6 must be discharged between recording level settings if any alteration is made to the presets or quite misleading results can be obtained.

On the foregoing machine, with an input of 2 mV at $1 \mathrm{kc} / \mathrm{s}$ to the microphone socket, and a valve voltmeter measuring the output at the anode of the stage subsequent to the controlled valve, the control preset must first be turned fully clockwise and then the record level preset turned slowly for a reading of 12 V . Then the input is increased to 20 mV and the control preset adjusted until this same reading is obtained.

If one is too ham-fisted, and the reading passes the necessary level, one must discharge C6 and start


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 TWIN GANG．＂0， $29208 \mathrm{pF}+176 \mathrm{pF}^{2}, 10 / 6 ; 365 \mathrm{pF}$, minia－
trie $10 / ; 500 \mathrm{pF}$ ，standard with trimmers， $9 /-$ ；midget， $7 / 6$ ； midget with trimmers， $9 /-500 \mathrm{pF}$ ．slow motion，standard $9 /-$ Small 3－gang 500 pFF ． $18 / 6$ ．Single＂ 0 ＂ 365 pF ． $7 / 6$ ．
 TUNING．Solid dielectric． $100 \mathrm{pF} ., 300 \mathrm{pF},, 500 \mathrm{pF}, 3 / 6$ each TRIMMERS．Compression ceramic 30,50 pF．， $3 / 6$ each $100 \mathrm{pF} ., 150 \mathrm{pF} ., 1 / 8 ; 250 \mathrm{pF} ., \mathrm{I} / 6 ; 600 \mathrm{pF}, \mathrm{y} 70 \mathrm{pF} ., 1 / 9$.

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SMALL， $300-0+300 \mathrm{v}, 70 \mathrm{~mA}, 6,3$ ． m .4 a ．
HEATER TRANS． 6.3 v．11 $\frac{1}{2}$ a．， $7 / 6 ; 6.3 \mathrm{v} .4 \mathrm{a}$ ．
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all over again. As a check, increase the input to 100 mV , when readings should be between 6 and 24 V ; reducing the input to 2 mV and discharging C6 again should then give a reading of 8 V or more.

## Constant Record Level

Enough has been said to have emphasised the difference in automatic gain control circuits as employed in tape recording from the simpler circuits of radio and television receivers. The difficulty of ensuring a full dynamic range, even when fortissimo and pianissimo passages alternate in quick succession, requires quite stringent design limits and more care in setting up these machines.

Our final example, perhaps a classic of its kind, is the Grundig TK18, which, though not the first, was perhaps the most widely publicised when it was launched in 1963. The essentials of this circuit are shown in Fig. 9.
part of a potential divider across the h.t., to give better regulation of cathode bias voltage.

Because of this the cathode by-pass capacitor is rated at 70/80 volts working-a small point but one which may save a small explosion if it is overlooked!

The second triode acts as a cathode follower, and the signal is passed to its grid, while the lower end of the grid load returns to a tapping in the cathode network. This brings the bias point to the straight portion of the valve characteristic.
$\therefore$ The rectifier MR2 handles the audio signal and charges the $10 \mu \mathrm{~F}$ capacitor C6. This is an MKT type, and not an electrolytic, and has to have a working voltage of 125 volts. The bias is fed to the grids of both the EF86 input amplifier and the vari-mu controlled valve, EF83.

The cathode of this valve is biased via the recording sensitivity control in the cathode of the first half of the ECC81. The shorter period of time constant during microphone recording, where staccato speech signals may be encountered, is


Fig. 9: Control circuit of Grundig TK18, with only the essential parts of circuit included.

The important points to note are that the delay period is as much as 15 minutes for a high level (radio or pick-up input) signal and 3 minutes for the microphone signals. The machine runs at maximum recording level until a strong input is applied, when the bias takes over.

The reason for the auto-gain control not coming in before a small signal is applied is to reduce the hum and noise at high level.

In this case, the control signal is tapped off from the same point as the feed to the recording head, via a $68 \mathrm{k} \Omega$ and $0.01 \mu \mathrm{~F}$ capacitor in series, and applied to the grid of the second half of an ECC81 double triode.

In the cathode of this valve there are two preset resistors, the Threshold VR2 and the Recording Sensitivity VR1 controls. These provide an overall setting of bias so that signals of more than 10 volts will be amplified, but not those beneath this level. It will be noted that the cathode circuit is actually
provided by switching a $6.8 \mathrm{M} \Omega$ resistor, R12, across the leak circuit, by S 2 , the microphone switch.

AT1 and AT2 are parts of the Record and Start switch contacts and serve to mute the circuit when the machine is switched to neutral.

It can be seen that the setting of the Threshold control is vital for correct operation, and, as mentioned previously, care must be taken when adjusting not to over-run the right point.

Space limitation prevents a detailed description of the setting up procedure, which requires an audio generator, a valve-voltmeter, and various networks of resistors across which the readings have to be made. Readers who may require this information can obtain it through the Enquiry Service.

The prime purpose of this article has been to show the methods and techniques of control used in tape recording, and, while by no means exhausting the range, may have demonstrated that this is not such a simple matter as it may at first appear.


THE variable power supply unit described here has been knocked-up by the author from his "bits box" to provide a d.c. source to help him when experimenting with transistor circuits. Meters for monitoring the output voltage and current are included in this transistorised unit, which has a continuously variable output from zero to 20 V d.c. and will give about an amp on the lower voltages.

From the circuit it can be seen that the voltage control is achieved by varying the base potential of the series power transistor Tr1, which operates as a current limiting device. As the base/ collector potential of this transistor falls, more current is allowed to pass and thus, the output voltage increases.

The base potential of the series transistor $\operatorname{Tr} 1$ is varied by the other transistor in the circuit, which can be considered as a variable resistor across the collector and base of the series power transistor Tr1. As the control transistor $\operatorname{Tr} 2$ conducts (its base/collector potential being reduced by the potentiometer in the base circuit), the d.c. resistance path from emitter to collector falls and at maximum conduction the emitter voltage is almost the same as that on the collector. Thus the collector/base potential of the series power transistor $\operatorname{Tr} 1$ is reduced to near zero, which allows maximum current to fow through it.
The current required for the control transistor is quite small. In fact it is so small, the author's unit utilises signal diodes (OA85) to rectify the output of a heater transformer. This part of the circuit makes use of a split capacitor arrangement to voltage double the output of the transformer

By H. Wagner

before it is applied to the voltage adjusting potentiometer in the base of the control transistor.

The shunt resistor value has not been included as this will vary with the meter to be used, which should have a full scale defiection of 100 mA . The easiest way to get the correct resistor is to put 100 mA through the meter and then adjust the length of the shunt resistor wire until a reading of 10 mA is obtained.

Precise construction details have not been included in this article as almost any type of meter and transformer can be used so long as the electrical specifications are similar to those given on the circuit diagram.

Some of the components used by the author can be substituted. For example, the smoothing choke in the control circuit supply could easily be replaced by a $470 \Omega \frac{1}{2} \mathrm{~W}$ resistor.

As far as the metalwork is concerned, almost anything can be used. The author built his unit in a tin box that previously contained cream crackers. The lid was replaced by a piece of paxolin, which supports the meters, switches and some of the components.

Layout is not critical, but it should be remembered that the power transistors, especially the series transistor Trl, should be mounted on a heat sink. Mica washers should be used when attaching the transistors to the heat sink which can be made from 16 s.w.g. aluminium sheet.

No protection facilities are provided on this unit, so it is important to avoid accidental shorting of the output which can result in permanently damaging the series power transistor Tr1.



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# practically Wireless commentary by IEINII 

NTOT even the most gullible reader imagines this article was dashed off while the toast cooled this morning. Scenes of a copy-boy straining at the leash while the "star reporter" tapped out his world-shaking scoop are commonly regarded as fiction. Even the Night Editor's office of the great

daily newspaper is less hectic than some playlets have made it appear.

When Henry announces an item of news, he has to face the disturbing possibility that in the meantime some clot will have superseded it, invented an alternative, elected a new Government or declared a revolution before the tit-bit appears in cold print.

This is partly the cost of electronic development. In one trade paper (mid-March) a correspondent recalled that in his day as an apprentice at an electrical accessories factory he was used to his foreman saying: "In five years, what we are making will be quite out of date."
Nowadays, in the electronics and communications sector, five days seems to be more like it. Some call it progress. This correspondent is often tickled by the way apparent "breakthroughs" are based on the ideas of our illustrious predecessors. And not only in theory.

Hence the commentary last February, which mentioned a few of these curious throwbacks. Hence, also, this long delay in answering
a criticism of that article, by John Niven Douglas, which appeared in the correspondence section of the April Practical Wireless. Between the first fine careless rapture of inspiration and the more careful perspiration that accompanies publication, there is a timelag.

This leads to a further shameful situation: the references that your columnist used may have been committed to that dusty limbo in the darkest corner of the workshop, or the magazines from which the reports were culled.

Since Mr. Douglas' polite reprimand was received, Henry has feverishly ransacked his back copies, but failed to turn up the news item about "mechanised h.f. systems". But this does not alter the basic argument.

Frequency synthesis, as Mr Douglas rightly points out, provides a system which "combines the stability of a crystal oscillator with the flexibility of a v.f.o.". It is indeed true that the frequency synthesiser is used by short-wave broadcast stations as the r.f. drive input to the transmitter. But this does not preclude its more sophisticated employment, even though attempts to limit the high cost have resulted in inferior units.

It is a fact that the principle has been allied to modular concepts and logic circuits to produce a system of lower cost and a high order of output purity. The Wadley system and the phase-locked loop system have affected receiver design, and constant development, using micro-miniaturisation techniques and computer channel selection, is already enabling the "con-stant-tuning-constant-lock" communications device to lend itself attractively to defence departments.

By using a pulse system of reference signals, both transmitter and receiver can be continually changing frequency-but precisely in step. To an eavesdropper, the frequency changes seem almost at


Where's that quote?
random, and the whole point of the modern development is the speed of the switching and the maintenance of frequency accuracy.
No, Mr Douglas, while Henry, in his ignorance, may have got hold of a wrong stick or two, he has this time grasped it by the right end.

Getting hold of the wrong stick is inevitable in this business of preparing a column some weeks before it will join the bills and the billet doux through the morning letter-box. The pace of electronic progress is such that predictions are outdated before they have left the pen. Jules Verne would have had a whale of a time in the nine-teen-sixties.

Yet there are certain lines of progress which seem inevitable, awaiting only the refinement of existing techniques, the advent of new materials and the release of classified information.

As I write, a report lies before me of a flat TV tube suitable for use in portable colour receivers. This news comes, of all places, from California. Yet I seem to remember Dr Gabor, now Professor at the Imperial College, putting forward his idea for this device a number of years ago.
Is this another example of the British hovercraft "complex", or are there still hopes that we might secure a small corner in the furthering of our own inventions?

# THE 'IMPERIAL' 3-band transmitter F.G.RAYER G3OGR 

THE box for the v.f.o. is easily made from "universal chassis" sections. One section has its flanges cut so that it can be bent to form the front and inner side of the box, and the flat top plate is then bolted on. The v.f.o. is then completely wired. After wiring, bolt the halfcompleted box to the chassis. The v.f.o. and transmitter can be tested before enclosing the box. This is done by bolting on the outer side, then fixing on the small back plate with self-tapping screws.

To avoid difficulty in securing correct coverage, a


Fig. 6: Construction and wiring of the v.f.o.

Wearite PHF6 coil or equivalent is recommended. The smaller winding, or primary, must be taken off completely. Then carefully unwind 31 turns from the grid end of the secondary. The remaining winding is L1. No means of adjusting the inductance is provided, as it was felt this was best. Trimmer TC1 is fixed to a bracket and reached through a hole in the side of the v.f.o box, and allows sufficient adjustment of band coverage.

VCl should preferably have two bearings and must be free from wobble. A small surplus 2-gang capacitor was actually fitted and frequency control was very smooth and satisfactory. No temperature compensating capacitor is included, on the grounds that this may in fact only give disappointing results. The whole v.f.o. is subjected to little heat and drift from this cause is small, and much less than can possibly be read on the v.f.o. scale.

## TUNING ADJUSTMENTS

A quick check of v.f.o. coverage can be made by taking R2 to a convenient h.t. supply, and listening for the carrier with a receiver. Adjust TC1 to give coverage from $3 \cdot 5 \cdot 3 \cdot 8 \mathrm{Mc} / \mathrm{s}$, with a little to spare each end of the band.

When all construction is finished, and the v.f.o. box tightly bolted up, exact calibration can be undertaken. This is best done by using a crystal calibrator, or $100 \mathrm{kc} / \mathrm{s}$ crystal marker, in conjunction with the receiver. The $3 \cdot 5,3 \cdot 6,3 \cdot 7$ and $3 \cdot 8 \mathrm{Mc} / \mathrm{s}$ points can be marked, and the v.f.o. can be heterodyned against the crystal to note if there is any change in frequency with vibration, or drift. Neither of these should be troublesome nor very apparent.

To simplify calibration, the same v.f.o. band was used for the higher frequency bands. This means that $3.5 \mathrm{Mc} / \mathrm{s}$ on the v.f.o. is $7 \mathrm{Mc} / \mathrm{s}$ and $14 \mathrm{Mc} / \mathrm{s}$. With the v.f.o. tuned to $3.55 \mathrm{Mc} / \mathrm{s}$, the transmitter output frequency is $7 \cdot 1 \mathrm{Mc} / \mathrm{s}$ or $14 \cdot 2 \mathrm{Mc} / \mathrm{s}$, according to the band in use, and so on. That is, the v.f.o. scale
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is multiplied by 2 or 4 , for 40 and 20 m bands respectively. These other frequencies can be calibrated throughout by tuning the v.f.o. against a $100 \mathrm{kc} / \mathrm{s}$ crystal marker, noting that $3.5 \mathrm{Mc} / \mathrm{s}$ corresponds with $7 \mathrm{Mc} / \mathrm{s}$ and $14 \mathrm{Mc} / \mathrm{s}$, and so on.

If required, it is easy to open out the tuning on the h.f. bands, as described later. This has no actual effect on efficiency, but does simplify tuning. With a smooth ball drive and large knob, it may be felt this modification is not wanted.

## TUNING THE EXCITER

H.T. must on no account be applied to the p.a. until grid current is available, and an aerial or other load must be connected. With S1 at 20, adjust the core of L2 for resonance at about $7 \cdot 1 \mathrm{Mc} / \mathrm{s}$. Resonance may be found with a wavemeter, or by noting the grid current of V5, the net switch being closed. Also check that VC2 allows tuning L3 to 80, 40 and 20 m bands. Tuning here is for maximum grid current, as shown by the grid meter, but this should be kept down to 2.5 mA maximum, by adjusting VR1 as required. V5 is normally operated with about 2 mA grid current ( 44 V bias across R10).
To test the transmitter, a 60 watt household lamp is convenient as a load. A first test is best on 80 m . Set the exciter bandswitch to 80 and adjust VR1 for about 2 mA grid current, with VC2 tuned for maximum grid current. If desired, a reduced h.t. voltage can be applied to the p.a., or a $100 \mathrm{k} \Omega$ resistor may be temporarily connected in series with R11, to keep the input low. With VC4/VC5 closed, switch to "transmit" and quickly tune VC3 for minimum anode current. The input will be small, and is increased by opening VC4/VC5, at the same time always re-adjusting VC3 for minimum current. With normal screen voltage and a reasonably high anode potential, a 60 W lamp can be lit brilliantly.


With a high voltage, V5 may be destroyed in a few seconds, if operated without grid drive, or off resonance. In these conditions, the input is much higher than normal, and is dissipated as heat in the valve itself. Typical ratings for the 6146 are 112 mA anode current at 600 V , or an input of about 67 watts, screen current then being 9 mA at 150 V , with 2.8 mA grid current, and an output of 52 watts. The actual transmitter was, however, generally used with 100 mA input, and 2 to 2.5 mA grid current.

With high level modulation, results are satisfactory with a wide range of inputs and p.a. voltages. A 400 V supply allows 40 W input at 100 mA , and this has proved to be very satisfactory. Typical circuit point voltages are shown in Fig. 1. These were taken with a $2 \mathrm{k} \Omega / \mathrm{V}$ meter.

When it has been noted how the p.a. tunes, there is no need to employ the reduced h.t. voltage, or additional resistor in series with R11. However,


Fig. 7: Circuit diagram of the complete modulator.
rapid tuning of the p.a. is necessary. Tuning is more critical on 40 and 20 . On the higher frequencies, anode tuning results in some change of grid current (this is usual with an un-neutralised power amplifier). The convenience of having separate grid and anode meters permanently connected is then very apparent.

Sufficient grid drive was obtained with a 250 V supply to the exciter. The current drain is about 50 mA . The second supply will normally be 300 V or higher, to permit 30 W or more input. It should be able to deliver $200-250 \mathrm{~mA}$. The screen voltage of the 807 's should not exceed 300 V . This means that R 30 can be reduced with a 300 V or similar low supply, but should be increased with a 500 V or larger supply, so that the screen voltage is about $250-300 \mathrm{~V}$. With a 500 V supply and 270 ohms cathode bias resistor, the 807's are rated to provide 32 watts output, which will easily modulate the p.a. With a 400 V supply and 40 W input, only about 20 W will be required from the 807's. The modulator should not be operated with V5 withdrawn. Screening cans are used on all except the 6146 and 807's and regulator V2.

## OPERATING THE TRANSMITTER

Brief operating details may be welcomed by anyone who is using this as a first transmitter. With a lamp load as described, speech should sound strong and distinct in a receiver. The receiver RF gain will probably need to be turned well back, and the receiver aerial input sockets may be shorted to chassis.

For normal operating, take a 75 ohms co-axial lead from the transmitter receiver aerial socket, to the receiver. Receivers with a 75 ohms input impedance will give best results with a dipole or other matched aerial. Interrupt one lead from the speaker transformer secondary to speaker, and take connections from here to the speaker muting sockets.

A dipole aerial can be used for any one band. It is usually about 128 ft . for $80,66 \mathrm{ft}$. for 40 , or 33 ft . for 20. The transmitter should load without difficulty into such a dipole, on the band for which the dipole is intended. Occasionally, the position of the aerial may alter its frequency, so that its length has to be changed slightly.

A balanced system, using 75 ohms twin-lead from a balun or tuner, is less likely to cause TVI than a co-axial fed aerial. An end-fed aerial generally requires a tuner, such as an aerial may be operated on all bands. The transmitter has been operated without TVI on the home TV receiver, with both dipoles and end-fed aerials, on 80, 40 and 20. This does not mean it will necessarily be free from TVI. in other circumstances, where the run of feeders, or other details, may be unfavourable.

The transmitter net switch allows the v.f.o. to be tuned to a clear frequency or to a received signal. After tuning for grid current, switch to transmit, and adjust loading (and grid current, if needed). This switch then provides complete control.

If it is wished to save a little on building costs, the surplus 807 may be used for p.a. The screen grid should receive about 250 V and input is up to 100 mA , 600 V maximum. Grid current can run at about 3.5 mA . No other changes are required.
C.W. and h.f. bandspread details next month.

NEXT MONTH in


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NORTHERN HEIGHTS AMATEUR RADIO SOCIETY

THE Society was formed after a meeting of persons interested on April 9th 1961. We were fortunate in having available for our fortnightly Wednesday meetings at the Sportsman Inn, Ogden, Halifax, a room at reasonable rent which is always comfortable and has such a fire in winter that we have no qualms about leaving our own firesides to attend a meeting.

## Syllabus

A syllabus was arranged, an early item being a Junk Sale to help. raise funds for the purchase of a communications receiver for the use of a patient in a local Cheshire Home. Another early item on the agenda was the first of a series of our highly successful "Pea and Pie Suppers"-which are even more popular than the Annual Dinner-if food is involved, the turn-up is amazing! The Manchester Society have always been our guests on these occasions, the association between the two societies being the outcome of a meeting of a member of the Manchester Society, under highly respectable circumstances, with a lady member of our Society, on the sandhills at St. Annes-on-Sea. The true and highly respectable story of this meeting has now been lost in the mists of antiquity and still grows more juicy with every telling, especially on the Lancashire side of the Pennines!
D. Garlick and scouts at G3MVH, Jamboree station of Halifax Boy Scouts Association (1962), operated by members of the Northern Heights Amateur Radio Society. The scout sitting is now G3TQQ.

Sufficient members wishing to take the R.A.E. were available for us to ask for a course at the local Technical College and this has since been supplemented by a course in Morse Code on another evening.
We have had visits to innumerable places of interest, thanks to the efforts of our indefatiguable Secretary, including trips to radio and TV stations, TV studios, County C.I.D. H.Q., an atomic power station, radio, TV and audio equipment factories, Jodrell Bank, trade film shows and our annual marathon (overnight both ways) to the Radio Communications Exhibition in London.

## Society Lectures

We have had lectures, to name only a few, on subjects ranging from "Lightning" (complete with demonstration of 2 ft . spark of artificial lightning!), "Radio Astronomy," "Radio on Stamps," to "Fire Prevention," "Tape Recording," and our old friend, W1BB, on tape and slides on " 160 m . DX'ing"-which has subsequently been lent to a large number of other societies after three full houses at our own meeting rooms. This is to say nothing of the more usual selection of subjects heard at radio society meetings-lectures on antenna problems, TX construction, s.s.b., v.h.f., D/F test equipment, station layout and a most illuminating and entertaining evening on "How I became a radio amateur," given as a joint effort by three of our younger members and which really brought the house down. We have shown technical films; as well as slides of the Society's outside events-the incidents on some of these, especially Field Days, being too incredible to relate.

Every year we have operated the station G3MVH for the Scout Jam-boree-on-the-Air and this event has been more successful in Scout contacts every year.

Twelve months ago our Society was granted the callsign G2SU, this being the callsign of the late Matthew Eskdale, a highly respected founder member of the Society and longstanding friend of many members.
This callsign was first used by the Society on a demonstration station at a local fete-one of several we run
every year-a time of anxiety and heartache (to say nothing of all kinds of ache from the aerial rigging) for those responsible. Although we have now done the demonstrations many times on each site, every occasion produces its problems-dicey trees and dubious canvas on marquees to be climbed for aerial rigging, scaffolding supplied without clamps, officious officials, open wire P.A. equipment, the sewing machine demonstration ("absolutely free from all radio and TV interference, Madam"-'"How in-teresting-in that case, would you like to come and listen to your sewing machines?"), the adjacent stand demonstrating the product of a small-time TV manufacturer where we had to go and connect up his aerials correctly for him in order to save our own reputations, power supplies which do not materialise at the time promised, gales, rain, storms (contact replies that the hail on our tent is "so $5+9$ " that he cannot hear our voices above it), working on live overhead cables protected only by very wet shoes and damp polythene sheet on wet trestle tables on a sea of mud, power failures, voltage regulation so bad that the rest of the lights on the ground dim when we switch to transmit, blowouts, aerials carried awaywe've had the lot and yet have always been fortunate enough by some means to keep the station on the air. We have ail learnt a lot in the process and as one of the team remarked on one occasion. "It's been a ${ }^{* * * *}$ of
a day, but I wouldn't have missed it for the world". It is these events which give an insight to up-andcoming youngsters on what happens (or fails to happen) amongst those wires stowed away at the back of the shack-many of these lads are now licensed themselves and we have the keenest and happiest crew of younger members one could wish for, and this fact, coupled with the experience of older members, is the lifeblood of any society. May we all go forward together for many more happy years in "our" Society.



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Free constructional details of the recommended cabinet are readily available from us．
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The unit forms the drive system of the＇Minette＇enclosure，for details see separate leaffet．

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insulation in these couplers can badly upset the biasing of the output valves and cause distortion and overheating.

The driver stage has load resistors both in its anode and cathode circuits. This gives antiphase voltage swings to the control grids of the output valves, required for push-pull operation. The driver has less than unity gain (since amplifier stages. Contact bias (or grid current bias as it may be called) is provided on the driver valve by the grid resistor, which is possibly in the order of $10 \mathrm{M} \Omega$. Note that it is returned to cathode.

Unbalance in the output valves often causes trouble. It can arise from unmatched output valves or alteration in value of one of a matched pair of comthe output stage is correctly matched d.c.-wise, a voltmeter connected between the two anodes of the output valves should read almost zero volts. Some amplifiers have adjustable bias on one valve to secure such a balance. This is normally done by fitting a variable resistor in the cathode circuit of one of the output valves-separate cathode biasing resistors for each valve being used with
this arrangement. Excessive unbalance can cause a high content of harmonic this arrangement. Excessive unbalance can cause a high content of harmonic distortion.

Fig. 2


 gain of the amplifier at spot frequencies can be determined. This can be plotted against frequency to provide a diagrammatical illustration of the amplige
ponse curve, as shown in Fig.

To the input of the amplifier should be connected the aerial and to the output the diode detector, as shown in Fig. 3. If the amplifier is wideband, a jumble of several local or high-power transmissions will be heard in the 'phones if the amplifier is working, while if the amplifier is tunable, it should be possible to select a local transmission for the test.

A very powerful transmission may induce sufficient signal into a good aerial to cause some response with the aerial connected direct to the diode circuit point " A " of the detector (Fig. 3). If the signal is non-existent or very input of the amplifier, then one can be sure that the amplifier has gain, at least.

R6 also aids with the thermal compensation of the output pair of transistors, and although only of a relatively small
value (usually a matter of tens of ohms-or less) it does limit the emitter conditions and may save the transistors from damage.
Tr1 is the "driver"
stage, which is an ordinary voltage amplifier loaded at the collector by a transformer winding. The secondary of this transformer is centre-tapped and thus feeds antiphase secondary of the output transformer to the base or emitter circuit of an earlier stage. The biggest trouble in this type of circuit, then, is crossover distortion which is a function of stage balance and biasing, and tests should be made along the
lines discussed above in event of the trouble.
Fig. 9(b) gives a circuit of a valve push-pull output stage. These are nearly always arranged as class $A$ amplifiers, and the one shown adopts the so-called
$\dagger$

example

 input signal turned wo to avoid ove coil and/or value of the tuning capacitane output peaks, keeping the some sort of indication on the meter at the output, and then adjusting the





The detector could be connected as in Fig. 3 and, with the approximate number of turns on the ferrite rod for 1.w. response, the tuning adjusted until the programme is heard in the 'phones (note: to enhance the pick-up it may be necessary to couple loosely an external aerial to the rod by winding a few
turns of the external aerial wire round the rod).
If the programme tunes (or approaches tuning) at full capacitance of the tuning capacitor more turns are required on the aerial winding proper. If then there are too many turns on the coil. Slight alteration in aerial coil inductance is obtained by moving the coil along the rod. At dead centre, the inductance is at maximum, and it falls as the coil is moved either side of centre towards

## D.C. Tests

We shall consider valve and transistor circuits side-by-side in terms of testing as we have done so far. The first logical step to take is to check the supply voltage. In valve equipment we have the heater voltage as well as the high
tension (h.t.) voltage to consider. Heater voltage shows its presence by a glowing tension (h.t.) voltage to consider. Heater voltage shows its presence by a glowing
valve heater. The anode The anode supply of an r.f. amplifier is often fed through the winding of
a coil, while the screen is fed through a medium to high value resistor (some-
times direct from h.t. positive, depending on the nature of the circuit). If a coil is possible. Lack of screen grid voltage should lead (a) to a check of the
The emitter/base bias of (b) is set by R2 R3, and this time the signal input is through an electrolytic C1. A high value is necessary with transistor audio stages owing to the lower impedance at the input relative to valve control grid circuits. R1 contributes a little towards the biasing, but here serves mainly for thermal stabilisation (see booklet No. 2). The output signal is developed following stage. The polarity The polarity of the coupling electrolytic capacitors is important, for reversed
connection when replacing could lead to bias disturbance due to d.c. leakage. The insulation of these couplers can be tested as with valve circuits by noting any change in d.c. voltage across the emitter resistor R1 when the component is disconnected. When it is first connected again, a kick in emitter voltage is normal, owing to the charging current taken by the electrolytic. Negative feedback and reduced gain can also result from open-circuit or value reduction
of C 3 .
Transistor power amplifiers invariably feature a push-pull output stage, using either a pair of p-n-p transistors or one p-n-p and one n-p-n arranged in a complementary pair, coupled to the speaker through an electrolytic instead of the conventional transformer.

 ment of the d.c. conditions of the stage can be gleaned by measuring the voltage transistor. Thus, if an emitter resistor is employed in the circuit, a basic assess-re-checked, for it must be here where the trouble lies current is about normal, the dynamic parts of the circuit should be carefully If to the measured voltage divided by the value of resistance in thousands of ohms.



 may not have emission. This can be checked in situ with a voltmeter simply by
 feed resistor and (b) to a check of the insulation of the bypass capacitor to
chassis. If this is shorting there would be no screen volts and the short-circuit

 trouble. When C 1 is connected from the anode of a previous stage, poor can result from bias 1o/pue uo!niozsp peg -sıəy!jdure fia roj se 8u!!nseəu ృo \$1s!suoo 1 isai opduis $\forall$ 'puouod. or by a faulty comcaused by lack of con-
duction in the valve
 bypass capacitor [ruenis sl! to pue
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aspect, we can analyse the circuits almost exactly as described for r.f. amplifiers. The fundamental difference being in the nature of the signal.

Of course, we rarely come across tuned audio amplifiers, though these do exist. Most audio amplifiers that we shall deal with have a relatively fiat reshigh audio (bass and treble) with hi-fi amplifiers.

As we saw in the previous booklets, these amplifiers are essentially for signal voltage or signal power-voltage or power amplifiers. The voltage amplifier amplifier which effectively translates the signal into audio power for operating $\qquad$ the loudspeaker.

Voltage amplifiers are also integrated with equalising networks and tone controls, allowing the programme signals to be tallored as may be required by the nature of the programme source. The volume control is sometimes between
 A basic voltage amplifier is given in Fig. 8 for valve and transistor at (a) and (b) respectively. At (a) the stage is biased by the volts drop across the cathode resistor R1, the control grid being returned to the negative end of this resistor. Thus, the cathodo positive relative to the grid (making 1 .

The input signal is applied through Cl , the signal being developed across R3. The valve 'boosts' the signal and it is re-developed in amplified form
 Test 1 is obvious and will indicate the supply voltage to the stage. The
voltage at test 2 will depend on the resistance value of the collector load. In
 voltage. Test 1 checks the full supply voltage test 2 the collector voltage and transistors and (b) for n -p-n transistors. These circuits show three tests for make one or two tests to prove whether or not the transistor is working as it Once we have established that the junction voltages are reasonable, we can as zero conduction there is no internal short-circuit in the transistor, this trouble would also show A third possibility is failure of or both of the transistor jurstor. If impression of normal conduction and two, lack of base bias showing as lack Thus, we have two possibilities: one, a fault in the transistor giving the voltage drop across the emitter resistor. the matter of a few $\mu \mathrm{A}$. No ordinary meter would show this current in terms of collector current (and hence the current in the emitter resistor) will collapse in the emitter/base junction. If for some reason this current fails, then the

or common point) makes the oscilator suspect.



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 to detail tests of this kind within the small compass of this booklet, but interested
 in circuits of that nature. The actual waveform generated can be displayed






 full supply voltage will be indicated irrespective of how the transistor is conducting. If there is a resistor in the order of thousands of ohms also in circuit, on how much current is passing through the collector resistor-the greater the current (hence, the greater the conduction of the transistor), the greater the volts drop across the resistor and the smaller the collector voltage. The base voltage will be very small, often well below one volt. This is because the emitter/base junction is always in forward conduction (by a matter of 9

is not told by this test, but often a knowledge of whether the stage is oscillating or not is sufficient.

This technique can be extended to most oscillators, including local oscillators of radio sets, erase and bias oscillators of tape recorders and oscillators used for test purposes and so forth. ио!̣е





s.sə!!!|dur * $\boldsymbol{y}^{\prime}$ I here are positive, lack of operation of the r.f. amplifier should lead to a more
detailed check of the signal or dynamic conditions. Thus we can prove the d.c. conditions of the transistor, and if reactions in emitter voltage. pletely. This should-if the transistor is working-result in a substantial drop
 current to avoid the possibility of transistor damage. This can be achieved resistor to increase or decrease the base current. It is best to decrease the
 the meter connected as for the emitter test in Fig. 4(b)-reversed polarity Now, tests to prove the goodness of the transistor are possible by changing
the base current while observing the voltage across the emitter resistor. With measurements, a meter of 100,000 ohms/volt sensitivity is desirable. to avoid excessive shunting of the resistor on the low range. For base voltage voltmeter is essential. A meter of not less than 20,000 ohms/volt is necessary

 дәјаи su!peà моI 'әл!
link wire round the wire from the oscillator to the oscillator section of the tuning gang and several turns at the other end round the ferrite rod aerial.
It must be remembered, of course, that the local oscillator frequency is removed from the frequency (or wavelength) shown on the be the i.f. The frequency may, in fact, be equal to the incoming or tuned frequency plus (sometimes minus) the i.f.
As already intimated, a change in d.c. conditions occurs when an oscillating stage is heavily damped so that oscillations cease. This is reflected in terms of a change in anode current of a valve or collector current of a transistor.

If there is no cathode or emitter resistor (some oscillators may not have these components), the voltage change can be registered across an anode or collector resistor.
If neither of these connections is feasible, a low-reading current meter can (i) (ii) and (iii) at (a) for a valve oscillator
Now, the voltage or current should be carefully noted with the stage operating served when the oscillator tuned circuit is damped with a fairly large value capacitor. A 1 or $2 \mu \mathrm{~F}$ usually kills all signs of oscillation in most circuits.
are preset, adjustable either by trimmer or dust-iron core. If the d.c. conditions
 quency equal to the i.f. to arrange some means of detecting this (see, for instance, Fig. 3) and then to adjust the tuned circuits for maximum output.

Some i.f. stages have fed back to them as bias a potential derived from the detector or a.g.c. diode. Valve circuits have a negative bias that rises in valuepotental so the reduce the conductivity of the with rising signal amplitude. The d.c. conditions of these circuits should also be taken into account when analysing the stage from the d.c. aspect.

Lack of gain or reduced sensitivity of i.f. stages is sometimes caused by alteration in value of one or more of the fixed capacitors across the i.f. transformer windings. The " $Q$ " or goodness factor of the winding may also deteriorof time.

## Frequency Changers

 as an i.f. amplifier plus a local oscillator, the two functions happening in the the "difference frequency" being selected by the output tuned circuits (the i.f. transformers) for subsequent amplification.
Most enthusiasts, however, have a transistor portable covering the range
of frequencies covered by the suspect local oscillator, and this makes a good

## 

 they have, they will almost certainly know how to apply them for frequencychecks. handy for tests of this kind, but few enthusiasts have such instruments. If oscillator signal is being generated. Being able to obtaincy is also desirable. A wavemeter or grid-dip oscillator is

When the local oscillator is in doubt, the d.c. conditions should be checked. them here. circuits of the various frequency changers, and it is not intended to reproduce detail the with arrangements for injecting an oscillator signal of suitable amplitude into
the mixer along with the incoming signal. Oscillators are considered later.


 a mixer and a local oscillator. With this system the mixer stage is almost
identical to an i.f. amplifier, though there are one or two differences in d.c.



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[^4]:    Rumania: Radio Bucharest (P.O. Box 111, Bucharest) gives date and frequency details on its QSL. The 15001530 English transmission beamed to Asia is now on 15,250.

    Switzerland: Swiss Broadcasting Corporation (CH 3000, Bern 16). has reintroduced its evening English transmission for the U.K. It is from 1845-2015 on $9,665 / 7,110$. The morning transmission from 11451315 remains on $9,665 / 11,865$. Other English transmissions affected by recent schedule changes are 0115-0245 6,120/9,535/11,715/11,775; 0415-0545 9,535/11,715/11,775; $0700-08309,595 / 11,775 ; 0845-$ 1015 15,305/15,430/17,830; 1330-1500 11,855/15,305/ 15,395/17,830; 1515—1645 11,880/15,255/15,305/ 17,830.

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