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## illustrated

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## CONNECTING WIRE

P．V．1：Pright $\qquad$ 4／＝

## 3 VALVE AMPLIFIERS

Kit of new parts，comsisting charwis． mains and output tramsinners，talsers
 gain sanplifier with wharate hase and

MAINS TRANSFORMERS Jixerlant xertlent vuality dinarantern．


## 13 CHANNEL T．V．S

Table Models．Famous Makes．Abso－ lutery Complete．Thum whe now
 $1 \% \mathrm{n}$
£2．19．0 linin．$^{£ 4.19 .0}$
LEAK AMPLIFIER SNIPS
Due to a iorturate purchase we are able to make the fors，All Amplicers have line matching output Translormers（Transiormers to match these Amplitiers to $15 / 2$ Speakers can be supplied at $22 / 6$ each）．
TL／12 WATT（slightly used）
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Pamphonic 60 watt Ampliters．
complete with pre－imp ilitiors all have the added facility of volume and tone controls so that they can be used without a pre－amp．il very bigh gaiu is not required．）

| （equired．） | TAN\％ 1 － |
| :---: | :---: |
| BULK BARGAINS $\dagger$ | $1-1-1-14-1,81-; 1<k 12-2-4-1 \text {, }$ |
| 12 POTS．Popular values．\％K゙ in |  |
| meg．Umised，mixai，pre－ $4 / 6$ | voblt Thetaly ructosed standaral |
| ONDENSERS， 25 Mixel．Electroly | mounting，puwerimu v pulles．7／6 |
| Many poputar sizes．List Our price Vatue |  |
| 100 RESISTORS 6／6 | AM／FM RADIOS |
| Excellent．sizer i－3 wati． | Fantaxtic orler．${ }^{\text {a }}$ values plus |
| 100 CONDENSERS 9／6 | Matity athi morsh． |
| niature ceramic and silver Mioa |  |
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| 25 TAG STRIPS 4／＝ | ENQUIRIES WELCOMED |
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| 100 HI STABS 9／6 | TELEPHONE C．O．D．ORDERS |
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Ginalantred berturtikances Joy Make Tuated 250k wonhing

VALVE HOLDERS，17Th，6d，



| TURRET <br> Mont makes．your chmiste ir anaik－ ahbr，with valie 18 P ，less talvers <br>  |
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| 1 Vaive 6d．，2－15，1／－． | GUARANTEE | SURANCE．Hatigfaction |
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| OZ4 | 4／616876 | 2／－120p3 | 12／8 | E | 8／－1 | 9 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1A7GT | 8／36E7GT | $4 / 9$ 20P4 | 1\％／－EA50 | 1／3 FW4／500 | 7／－U24 | 12／6 |
| 1c5GT | 7／－8K89 | 5／－20P5 | 15／－EABC80 | 8／9 GT1C | 12／61 UT5 | 10／6 |
| 1 D 5 | 7f－6K8GT | 11／－25A6G | 8／－EAC91 | 4／－GZ32 | 7／61U28 | $8 / 8$ |
| 1D6 | 9／8｜6K25 | $81625 \mathrm{L6GT}$ | 7／9 EAF42 | 8／3 G234 | 11／6 U31 | \％－ |
| 1H5GT | 8／－6L1 | $9 / 625 \mathrm{Y5G}$ | 8／－EB34 | 1／3 HK90 | 9／6 U83 | － |
| 1 L 4 | 3／－8L6G | $7 / 625 \mathrm{Z4G}$ | $7 /-$ EB41 | 5／－HL41DD | 8／8｜U35 | 12／6 |
| 1LD5 | 4／3 6L18 | $7 / 925 \mathrm{Z5}$ | 8／－EB91 | $3 / 3$ HN309 | 19／－U37 | 261 － |
| 1LN5 | 4／8 6L19 | 11／6 25Z6G | 81－E EBC33 | 4／9，HVR2 | 9／－＇U50 | 4／9 |
| 1N5G2 | 8／916LD20 | 7／9 27SU | 17／6 Ebc41 | 7／91KT32 | 819 U52 | 4／9 |
| 1 R 5 | $5 / 616 \mathrm{~N} 7$ | $7 / 6$ 30C1 | $6 / 9$ EBC81 | 7／9 KT33C | 4／－U76 | $5 / 6$ |
| 105 | $5 / 3 / 6 \mathrm{P} 1$ | $9 / 630 \mathrm{c} 15$ | 11／6 EBF80 | 716 KT 36 | 14／－ 773 | 9／6 |
| 154 | $7 / 68 \mathrm{P} 25$ | $8 / 630 \mathrm{~F}$ | 718 EBF83 | $9 / 6$ KT44 | 8／－U107 | $2 / 6$ |
| 155 | 4／6｜6P28 | $9 / 930 \mathrm{FL} 1$ | 916 EBF89 | 719 KT 55 | 1\％／6 U191 | 11／6 |
| 174 | 3／－898G | $8 / 630 \mathrm{Ll}$ | $6 / 6$ EBL21 | 10／9 KT81 | 11／－U2 | 16 |
| 2D21 | 5／618Q7GT | $8 / 9130 \mathrm{~L} 15$ | $9 / 9{ }^{\text {E E C }}$ 2 2 | 4／91KT63 | ${ }^{4 / 6}$ | 19\％－ |
| 3 A 4 | 4／－6879 | 9／－30P4 | 9／6｜EC91 | 4／6 KTb ${ }^{\text {d }}$ | 13／6｜${ }^{1 / 301}$ | $12 \%$ |
| $3{ }^{\text {A }} 5$ | $8 / 9$ ESA | $5 / 9$ 30P12 | $7 / 6$ EC92 | ${ }^{8 / 6}{ }^{\text {KT76 }}$ | 8／6 U309 | 8／6 |
| 3D8 | $4 / 9$ ESC7 | $4 / 9$ 30P19 | 13／61ECC31 | $7 / 6 \mathrm{KT88}$ | 19／－ 1 －${ }^{\text {¢ }}$ | 18 |
| $3 \mathrm{Q4}$ | 7 l －6SG7 | $9 / 9$ 30PL1 | ${ }^{9 / 3}$ ECC32 | 4／－KTWb1 | ${ }_{819}^{5 / 9}$ | 11／6 |
| 354 | 5－6SH7 | 3／－30PL13 | $9 / 6$ ECC33 | 9／9 KTW69 | $5 / 6$ |  |
| 374 | ${ }^{6 / 6} 688$ | $5 / 8 / 3505$ | $8 / 6$ ECC34 | 9／－KTW63 | $5 / 9$ U4 |  |
| ${ }_{5 R 4}$ | $9 / 6{ }^{\text {8 }}$ 6S 7 | 5／－35L6GT | $81-$ ECC35 | 5／9 KTZ ${ }_{\text {／}}$ | $\begin{array}{r} 11 /- \\ 3 /- \\ \text { U801 } \end{array}$ | 17／6 |
| $5 \mathrm{5T4}$ | $8 /-6 \mathrm{GL}$ 7GT | 5／9 35W4 | ${ }_{5 / 8}^{8 /-E C C 40}$ | 9／8 L63 |  |  |
| 5U4G | $5 / 6$ GSN7GT | 4／9 35Z9GT | $5 / 8$ ECC81 | 6／B L L 152 | 6／6 UABC80 |  |
| ${ }^{\text {5V4G }}$ | \％／6 6SQ 7 | ${ }_{3 / 6}^{5 / 9} 3{ }_{41}^{3525 G T}$ |  |  | $\begin{aligned} & \text { Y/OAF42 } \\ & 10 / 6 \mid O B 41 \end{aligned}$ | \％ |
| 5Y3G | 4／9 6SS7 $5 / 6 \mid 604 \mathrm{Cr}$ | $3 / 8 / 841$ $9 / 4$ | 6／B，ECC83 <br> $6 / 8$ ECC84 | ${ }_{7 / 6}^{6 / 6}{ }_{\text {N }} \times 8$ | 131－UBC41 |  |
| 5 Y 4 | 9／6 6VBG | 4／6 $50 \mathrm{B5}$ | $7 / 9$ ECC85＇ | $7 / 6$ N108 | 13／－UBC81 | 19 |
| 524 | 9／－6V6GT | 8／9，50c5 | $8 / 6$ ECC88 | 11／6，N152 | $8 / 3$ |  |
| 5Z4G | $7 / 68 \times 4$ | 4／9 50L6GT | $7 / 6$ ECF80 | ${ }^{8 / 9}$ P41 | 3／6 |  |
| 5 E 4 GT | 9／6／6x59 | $5 /-53 \mathrm{KV}$ | 9／6 ECF82 | ${ }_{8 / 3} \mathrm{P} 61$ | 2／9 |  |
| 8／30L2 | 9／－8X5GT | 6／－61BT | 17／6 ECH21 | ${ }^{10 / 9} \mathrm{P}$ | $8 \%$ UC02 | 19 |
| 6A6 | 3／9 786 | $10 / 9$ 61SPT | 11／－ECH35 |  |  | 19 |
| 6A7 | $9 /-787$ | $7 / 9$ 62BT | 13／6 ECH42 | $8 / 6$ PC97 | 9／6 UCc85 | 析 |
| 8A8G | $7 / 9785$ | 7／9，75 | 5／6．ECH81 | \％／－PCCS4 | 6／8 GEF80 | $10 / 6$ |
| 6A8GT | 12／6 ${ }^{\text {cb }}$ | 7／6，78 | 5 －ECH83 | $8 / 8$ PCCB5 | $7 / 9 \mathrm{OCH21}$ |  |
| gact | $3 /-7 \mathrm{H} 7$ | ${ }^{7 / 3} 80$ | $5 / 6$ ECL80 | ${ }^{6 / 8}$ Pcecs |  |  |
| 6AG5 | $2 / 9757$ | 15／－83 | 9／6 ECL82 | 8／－PCCs8 | 9／6 ${ }^{\text {UCL } 82}$ | \％19 |
| ${ }^{\text {bag }}$ | 6／9 7Y4 | $81-1858 \mathrm{BT}$ | 19／6 ECL83 | 10／6 PCC18\％ | 121－UCL82 | 9\％ |
| 6AK5 | 5／－7Z4 | 8／－185BTA | 9／6 ECL84 | $121-\mathrm{PCC188}$ | 819 | $1 \cdot$ |
| 6AL5 | $3 / 38 \mathrm{D}$ | 3／－807 | ${ }^{7 / 6}$ ECL88 | $10 / 3$ PCF80 |  | 18 |
| ${ }^{6 A M 6}$ | $3 /-10 \mathrm{Cl}$ | 11／6813 | 491－EF22 | 7 T PCF884 | 12f－UF42 |  |
| 6AQ5 | 8 B －1002 | 14／6．832 | 14／－EF30 | $3 / 3$ PCFF88 |  | $3 /-$ |
| 6ATB | 5／－10F1 | 4／9 888A | 12／8 EF37A | $81 / \mathrm{PCL88}$ | 7／3 UP85 |  |
| 6AUB | \％／－10F9 | 1016954 | ${ }_{2 / 3} / 9$ EF39 | 716 PCL83 |  |  |
| 6avb | 5／9，10F18 | 10／－955 | $2 / 3$ EF40 | 11／－PCL83 | $7 / 3$ UL41 | 81 |
| 6B7 6886 | ${ }_{3 /-10 \mathrm{P} 12}^{10 \mathrm{LD} 11}$ | $14 / 6958$ $8 / 61625$ | 2／6－EF41 | 87－PCL84 |  |  |
| 6B8G | 3／－10P12 | 8／6 1825 | $5 / 6$ EF42 | $6 / 9$ PCLS8 | ${ }_{10 / 6}^{10 / 6}$ UL48 |  |
| ${ }_{\text {6BA6 }}^{686}$ | 5／6）10P14 5／6／10P18 | $9 / 6,5763$ $7 /-9001$ | $\begin{aligned} & 7 / 6 \\ & 3 / 62 \end{aligned}$ | ${ }_{1 / 6} /{ }_{\text {PCLS }}$ |  |  |
|  |  | 2／－9001 | 3／6 EF50（A） | $2 / 8$ PEN45 | 3／8 UM80 UR1C | ／9 |
| BBH6 | 6／6 12AH8 | 11／－9003 | 5／9 EF54 | 3／3 PEN46 | 4／6 บบ6 |  |
| 6BJ6 | $5 / 912 \mathrm{AT} 6$ | 8／8 ATP4 | $2 / 6$ EF80 | 4／6 PL33 | $9 / 6$ UU7 | 9／B |
| 6BR7 | 8／6｜12AT7 | 5／6 A Z31 | \％／6 EF85 | 6／－PL36 | $9 / 6$ UV8 | 13／6 |
| 6BR8 | $9 / 6$ 12AU6 | 9／－AZ41 | $7 /$. EF88 | $7 / 6 \mid$ PL38 | 15／－LYIN | 11／－ |
| 6BW6 | $7 / 912 \mathrm{~A}{ }^{\text {c }} 7$ | 6／－1836 | 6／9 EF89 | 6／9，PL8L | $8 / 3 \mathrm{CY} 21$ | $9 / 6$ |
| 8BW7 | $5 /-12 \mathrm{avb}$ | $6 / 9 \mid$ cle | 8／－EF91 | 3／－PL82 | 6／8 UY41 | $6 \%$ |
| ${ }^{6} \mathbf{C 4}$ | $2 / 312 \mathrm{~A} \times 7$ | 6／6 CBL31 | 13／6｜EF92 | 3／－PL83 | ${ }^{6 / 6}$ UY85 |  |
| 6C5 | $5 / 612 \mathrm{BA} 6$ | $7 /$－ CCH 35 | 13／6 EF183 | 9／9，PL84 | \％／－FP4B | 9／－ |
| 6c6 | 3／6，12BE6 | $8 / 6 \mathrm{CL} 33$ | 91－EF184 | 9／61PL320 | $8 / 3$ VP23 | $2 / 9$ |
| 6C9 | 11－12BE5 7 | 819，CY31 | \％／6 EK32 | 716 PM84 | 8／－VP41 | 5／8 |
| 60D6G | 19／612088 | $5 / 6 \mathrm{D7}$ | 3／3ELL32 | $3 / 9, \mathrm{PX4}$ | 12／6 VR105 | 5／6 |
| ${ }^{\text {62 }}$ 6 | ${ }^{8 / 6} 12 \mathrm{E} 1$ | 1／／6 DA30 | 11／6｜EL33 | $11 /-\mathrm{PX} 25$ | $9 /$－｜VR150 | $5 \%$ |
| ${ }^{612} 2$ | $3 / 312 \mathrm{H6}$ | $1 / 9$ DAC32 | 8／－EL34 | 11／6 PY31 | \＄1\％W76 | 4／9 |
| 6D3 | $9 / 61255 \mathrm{aT}$ | 3／3 DaF91 | $4 / 6$ EL35 | 6／－PY32 | 10／－｜W81 | 7／3 |
| 6D8 | 3／－12579T | 8／－DAF96 | 713 EL38 | 12／6 ${ }^{\text {P Y }}$ 33 | $10 / 6 \mathrm{~F} 61 \mathrm{M}$ |  |
| ${ }_{6}^{6 F 1}$ | 4／9 12K7GT | $4 / 9$ DF33 | $8 / 8$ EL41 | $8 /-\mathrm{PY} 80$ |  |  |
| $6 \mathrm{F6}$ | 7／612K8 | 9／9 DF91 | 3／－EL42 | $9 / 6$ PYS1 | ${ }^{6 / 3} \times 65$ | $81 \%$ |
| ${ }^{\text {6F6G }}$ | $4 / 612 \mathrm{KgGT}$ | $11 / 6{ }^{\text {DF96 }}$ | $7 / 3$ EL81 | $8 / 9$ PY89 |  |  |
| ${ }^{6 F 13}$ | 4／9｜2Q：GT | 4／9 DF97 | 7／6，EL84 | ${ }_{9 / 6}^{6 / 6}$ PY83 | $6 / 918$ | 11／－ |
| ${ }_{8}^{6 F 14}$ | $9 / 61887$ | $7 /-\mathrm{DHB} 3$ | $5 / 6$ ELP5 | $9 / 9$ PY88 |  |  |
| ${ }_{6}^{6 F 15}$ | $9 / 6125 G 7$ | $4 / 6 \mathrm{DH} 76$ | 4／6 EL91 | ${ }_{6 / 9} 9$ PY800 | $7 / 9 \times 78$ | 21／\％ |
| ${ }_{6}^{6 F 38}$ | $8 /-12 \mathrm{SH}^{6}$ | $3 / 6 \mathrm{DK} 32$ | ${ }_{5 / 6} 9 / 6$ EL95 |  |  | 91. |
| ${ }^{6 F 32}$ | $4 / 9$ 12S57 | 5／6 DK91 | $5 / 6$ EM34 | $8 / 9$ PZ30 | $9 / 6 \times 83$ | $8 /$ |
| ${ }^{6533}$ | 4／－12567 | $4 / 6$ DK92 | 77. EM80 | ${ }_{8 / 616}{ }^{\text {R19 }}$ | $9 / 6 \mathrm{Z} 63$ | $4 / 8$ |
| ${ }_{6}^{645}$ | ${ }_{\text {1／6 }}^{1 / 3125557 \mathrm{GT}}$ |  | /19\|EM81 |  | 11／－ $\mathrm{Z}_{66}$ | $8 / 6$ |
| ${ }_{6}{ }^{\text {J } 50} 6$ | 3／－1303 | $8_{6 / 6}$ DL 35 | $2 / 6$ EM85 | ${ }_{9 / 6} \mathrm{SPP}^{\text {S }}$ | $2 / 3$ |  |
| BJ5GT | $4 / 31487$ | 19／919L63 | 9／－EN31 | 181－SP81 | 100＇s |  |
| 8J6 | 3／6｜19AQ5 | 7／9 DL75 | 6／－EY51 | 2／6 SU25 |  |  |
| 8J7 | $8 / 6$ L9BG6 | 14／－DL82 | 9／－EY86 | $7 / 3$ SU2150 | 4／6 NO |  |
| ${ }_{60} \mathbf{7} 7 \mathrm{G}$ | 4／9 20Dl | $8 / 9$ DL92 | 5／－EY88 | $9 / 6$ T 41 | NOT |  |
| 6J\％GT | 7／8 20 FS | 9／6 DL94 | 8／6 EZ40 | $7 /-$ TDD4 |  |  |
| 6KGGT | 6／－20L1 | 16／－DL96 | $7 / 3$ EZ41 | 7／－ 014 |  |  |
| 6K7 | $5 / 9$ 20P1 | 9／8 DM70 | 5／9 EZ80 | $5 / 9 \mathrm{U} 18$ | ENQS． |  |



CLEAR plastic PANEL meters First grade quality. Cofl pane meters. avallable ex-stock S.A.E. for illustrated leafiet. Discounts for quantity. Aval:able as follows. Type MR, 38P. ${ }^{111 / 24} 12$. square fronts
 $500 \mu \mathrm{~A}$ A $100-0-100 \mu \mathrm{~A}$ $100-1$
$\operatorname{lom}_{5} A$
5 MA
10 mA
10 mA
50 mA
100 mA
100 mA
200 maA
300 mA
$\begin{array}{lll} & 20 / 6 & 150 \mathrm{~V} . \mathrm{AC} 22 / 6 \\ 500 \mathrm{~mA} & 22 / 6 & 300 \mathrm{~V} . \mathrm{AC} 22 / 6 \\ & 22 / 6 & 500 \mathrm{~V} . \mathrm{AC} 22 / 8\end{array}$
TyMe MR.
$50 \mu \mathrm{~A}$. ${ }^{5216}$
5716 mma

$\begin{array}{llll} & 47 / 6 & 10 \mathrm{~mA} & 32 / 6 \\ 100 \mu A & 376\end{array}$ $\begin{array}{llll}500 \mu \mathrm{LA} & 37 / 6 & 100 \mathrm{~mA} & 32 / 6\end{array}$ b0-0-50 $12 \mathrm{~A} \quad 576$ 1A. DC $32 / 6$ $100-0-100 \mu \mathrm{~A} 47 / 6300 \mathrm{~V} . \mathrm{DC} 32 / 6$ Type MK. 300 F . AC $32 / 6$. B in. fronts. $50 \mu \mathrm{~A}$ 5016 min $100 \mu \mathrm{~A} \quad 59 / 61 \mathrm{~mA}$ 500491610 mA 5012A $39 / 6 \quad 100 \mathrm{~mA}$ |  | $59 / 6$ | $1 \mathrm{~A} . \mathrm{DC}$ |
| :--- | :--- | :--- |
| $100-0-100 \mu \mathrm{LA}$ | $39 / 6$ | $300 \mathrm{~V} /-$ |


POST EXTRA


HE-40 DE LUXE 4-BAND COMMUNICATION RECEIVER Frequency coverage $550 \mathrm{Kc} /=$ to $30 \mathrm{Mc} / \mathrm{s}$ con tinuous. Operathon $220 / 240$ volt A.C./J.C. The perfect recelver for short
wave listening. Special wave listening. Sildecial tuning dial electrical bandtuning dial-electrical band-Spread- 0 - 100 logging scaleimproved selectivity-mbilt limiter-B.F.O.-Phone jack bullt in 5 in . speaker-tonebult in sin. speaker-tone-control-sLaridby switch-supplied with threc aerlals, ferrite toop for broadcast band, adjustable 581n. Whip ior short wave and wire and guaranteed with manual. f24.15.0 carrlage paid. S.A.E. for full details. Part exchanges welcome.

MODEL ITH-2 MULTITESTER 20000 O.P.V.
20.000 o.p.v. D.C.: 10,000 O.p.V. A.C
$\quad \begin{aligned} & \text { cher } \\ & 0-6-25-250-600-2500 \\ & 0-10-50-500-1000 ~ V\end{aligned}$
$0.01-0.3 \mathrm{mFd}$ (at AC 10 V )

## $0-10-50-500-1000 \mathrm{~V}$ $0-50 \mu \mathrm{~A}-25 \mathrm{~mA}-250 \mathrm{~mA}$ $0-60 \mathrm{~K}-6 \mathrm{M} \Omega$ $-20 \mathrm{~dB}-+22 \mathrm{~dB}$ $0-10-50-500-1000 \mathrm{~V}$ $\frac{1.5}{31} \times 4 \mathrm{U} \times 12 \mathrm{in}$. <br> D.C. V: <br> Current: <br> Resistance. <br> Capacity. <br> Decobei: <br> Output rance. <br> Battery:

Supplied complete with batteries. leads instructions. 82/6. Post $2 / 6$.
MARCONI TFIA4G/4 STANDARD SIGNAL GENERATORS
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（＇ODIPIE＇IE，PONHER PACK HIT，1日／11 Consisting of Mains Trans．，Metal Rect1－ ther，Dulible electroty tic，simoothing choke chansis and circuit．For 200－250 V．A．C． madis．Output 250 v．， 60 mA ．E 3 v．， 2 a． IR．W．C．POWVH：IK PAC＇K．39／9．Louvred metal rase only $8 \times 5 \frac{1}{3} \times 2 t$ fri．Stove enamelled．For $200-250$ V．A．t．malns． Gutput at 4 pin plus and socket 250 V ． 60 mA ．fully smoothed and 6.3 v 2 a ．Sult－ able tor powet requirements of aimost
any Pre－amp．or Fadio Tuner any Pre－amp．or Fadio Tuner．
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| 32 High St | 6 | 13 Exchange S | 51 | 73 | 56 Morley St． | 8－10 Brown St． | －7 County |
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|  | Snow | ．）Shemeld |  |  | eatre）Br | Manchester 2 | s |
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HIGH FIDELITY AMPLIFIER AIO A highly sensitive Push－Pull high output unit with self－contained Pre－amp．Tone Control Stages．Certifled pertormance
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Carr $10 \%$ Instructions．II required perforated cover with carrs－ for 19／9．The amplifier can be supplifed．Tac－ ory built，with EL34 output valves and S．A．E．for leafet． TERMS：DEPO
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SUPERHET FEEDER EVIT．Design of a high quality Radio Tuner ispecially sultable for use with our Amplifiers）．
Delayed A．V．／C．
Controls are Tuning W／Ch．and Vol．Only $250 \mathrm{v} .15 \mathrm{~mA} . \mathrm{H} . \mathrm{T}$ ，and L ．T．of 6.3 v .1 amp ．required from ampliffer． Size approx． $9 \times 6 \times 7 \mathrm{in}$ ．high．Simple align－ ment procedure．Point－to－Point wiring diagrams，Instructions and priced parts list with illustrations，2／6．Total bullding cost

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 dit＇r．batteries and I．．T．\＆ ucromulitors when ron－ nectad to A．C．mains supply $200 / 500 \mathrm{v}, 50 \mathrm{ch} .5 \mathrm{STH} A 1 \mathrm{LH}$ FOIK AIL BATTEIGE RE： CiIVEFS normally using 2 v ，accumulators．Complete
ith daigrams and instructions． kit of parts with daigrams an
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Amplifier， 24.12 .6 ． 2 in ．R．A． 3 ohms iu watts （12．000 Lines）．5819．
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 dWALITY ADIPINIERS Sultable tor any record playing unit and most micro－ pass and Negative feed－back 12 dB ．Sepatate 250 v ． $50 \mathrm{c} / \mathrm{s}$ ．Output for $2-3 \mathrm{ohm}$ speaker． Mullard valves EZ 80 ，ECCB3，ELB4．Size only $7 \times 5 \times 511 n$ ．high．Guaranteed 12 months． Only 85.19 .6 ．Send S．A．E．Tor＇leatlet． Terms：Deposit $22 / 6$ and 5 monthly pay－ ments of $2 R / 8$ ．

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$18 \times 18 \times 10 \mathrm{in}$ ．Finish as above．Terms． Deposit $17 / 9$ and 9 monthly payments of 17／9．Only £7．19．6．Carr．8／6．
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A highlyesconsitive f－valye infality nmplifier for the homit．small eluls，atc．Only 20 millivoits input is ree quired for full output so that it is silitabie ror ase svith all orre Bebarate base and reble controfs are prowided．Theue sepadate jasa and trente controts are provided．Inese

 available tor the supply of a ikadje Feeder cinit．or l＇mile－Deck preamplifier．For A．c．maine input of 200 － 230－950v． $50 \mathrm{c} / \mathrm{s}$ ．Output for＇ $2-3$ ohms speaker．（＇hassis is not alve．Kit is complete in every detail and includes lnily puncheat ehassis（whin basephate）with Blue fammer lifish and moint－（o－point wiring diapramm and instruations．Exerptlonal


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polished walnut veneer．Matching im－ polished walnut veneer．
pedance 15 ohms．Frequenc
pecance 15 ohms．Frequency range $40-15.000$ cap．s．Power
handling 6 watts nominal．

8 Gns． Ideal for Stereo

Mt CTI－MEITEIR（AHE M1．Sensitivity 2，000 ohms per volt．A．C．and D．C．，54／－． A．10．Baslc Meter sensitivity 155 micro－ am ${ }^{\circ} \mathrm{S}$ A．C．and D．C．ranges 24.11 ．8．I．+0. Sunsitivity up to 10,000 ohms per volt A．C． and D．C．．\＆6．2．6． 30,000 ohms per volt．
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12in．for 8 or 18 x x． speakers，£4．11．9． GINIOIC MODEL Size $30 \times 20 \times 15 i n$ ． Suitable $\begin{aligned} & \text { Speaker．} \\ & \text { Speaker }\end{aligned}$ systems below
Only 7 ghs．
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 Incorporating the latest Collarg Studio Tape Transcriptor．The Audiotine likh Quality Tape Amplifier with negative leedback equalisation for each of speeds．High Flux P．M．Speaker．empty Tape Spool．aHepl of Best Quality Tape and a Handsome Portable Carrymg Cabinet tastriully covered in con－ Hew of Best Quality Tape and a Handsome Portable Carrymg Cabinet tastriully covered in con－
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Two input sockets with associated controls allow mixing of＂mike＂and gram．．as in A10 ECC83．Ei 84，EL84 ER81 Hieh Quality sec ECC83．EL84．EL84，EZ81．High Quality sec－ tionally wound output transformer specially designed for Ultia Linear operation and reli－ able small condensers of current manurac－ AND TREBLE＂LIft＂and＂Cut＂．Frequency AND TREBLE LIt and cut Frequency
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HATVERM CHAKGIKR KIRS Consistimz of Maics Trans－ ormer，well Bridge，Metal Rectifler，well ventllated steel Gasemmen parels Heavs Duty Arommets，parels，Heavs Duty tiv．or 12 z． 1 amp．．．．．．．．．．22／9 As above，with A mmeter，． $28 / 9$ 6 v． 2 amps．．．．．．．．．．．．．．． $19 / 9$ GV．or 12 v ． 2 amps．inclu sive ol Ammeter ．．．．．．．．．．．35／9 Ammeter and variable charge rate selector．．．．．．．52／9 cHIIEGEIR AMMEIEIRS 0－1．5 a．．0－4 a．． $0-7$ a． 819 ea． CHAIS（ililk FIT＇，I？s， 10 Amot． Whth variable charge rate adiustment and amm
£4．19．6．Carr． $10 /$ ．

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Interleaved and Impregnated，Frim－FLIA，NHikotimin（eontinued）－
 $250-0-250 \mathrm{v}$ ． $70 \mathrm{~mA}, 6.3 \mathrm{v}, 2 \mathrm{a}, 0-5-6.3 \mathrm{v}$ ．2a $350-0-350 \mathrm{v} .80 \mathrm{~mA}, 6.3 \mathrm{v}, 2 \mathrm{a}, 0-5-6.3 \mathrm{v}, 2 \mathrm{a}$ $250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v}, 2 \mathrm{a}, 6.3 \mathrm{v} .1 \mathrm{a}$
$250-0-250 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v} .3 .5 \mathrm{a}, \mathrm{C} . \mathrm{T}$. $250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .3 .5 \mathrm{a}, \mathrm{C}, \mathrm{T}$
$250-0-250 \mathrm{v} .100 \mathrm{~mA}, 6 \mathrm{v} .4 \mathrm{a}, 0-5-6.3 \mathrm{v}$ $300-0-300 \mathrm{v} .130 \mathrm{~mA}$ ． 6.3 v ． $4 \mathrm{a}, 6.3 \mathrm{v}$ ．1a，for Mullard 510 Amplifer
$300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a},{ }^{*} \dot{0}-5-6 . \ddot{3} \mathrm{v} .3 \mathrm{a}$ $350-0-350 \mathrm{v} 100 \mathrm{~mA} .3 \mathrm{v} 4 \mathrm{a}$ ， $5-6 \mathrm{v} .3 \mathrm{a}$ $350-0-350 \mathrm{v} 150 \mathrm{~mA}, 6 \mathrm{v} 4 \mathrm{O}-5-6 \mathrm{v} 3$ 3 $26 / 9$ FULIV：SHROLIMEIS UFIRIGIIT $250-0-250 \mathrm{v}, 60 \mathrm{~mA}, 6.3 \mathrm{v} .2 \mathrm{a}, 0-5-6.3 \mathrm{v}, 2 \mathrm{a}$ Midget type 2k x $3 \times 3$ अin． $250-0-250 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 0-5-6 \mathrm{v}, 3 \mathrm{a} 1711$ $300-0-300 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a} .0-5-6.3 \mathrm{v}$ ． 3 a 27／11 $300-0-300 \mathrm{v} .130 \mathrm{~mA} .6 .3 \mathrm{v} .4 \mathrm{a}$
1 a ．for Mullard Amplifer $350-0-350 \mathrm{v} 100 \mathrm{~mA}, 3 \mathrm{v} 4 \mathrm{a}, 0-63 \mathrm{v} 3 \mathrm{a} 33 / 9$ $350-0-350 \mathrm{v}, 100 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 0-5-6.3 \mathrm{v} .3 \mathrm{a} 2711$
$425-0-42 J v, 200 \mathrm{~mA}, 6.3 \mathrm{v} .4 \mathrm{a}, \mathrm{C} . \mathrm{T} .5 \mathrm{~V} .3 \mathrm{at}$.
$425-0-425 \mathrm{v} .200 \mathrm{~mA}, ~ 6.3 v . ~ 4 a . ~ С . Т ., ~$
6.3 v
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Push－Pull 8 watts，ELB4．or 6VG＇to 3 ＇s or matched to $15 \Omega$
Push－Pull 10－12 watts to mateh 6 V 6 or EL84 to 3－5－8 to $15 \Omega$
Following types for 3 and $15 \Omega$ speakers Push－Pull $10-12$ watts $6 V 6$ or FLS 84 Push－Pull Mullard 510 Uitra Iinear Push－Pull 20 watts，sectionally wound 6L6．KT66，EL34．etc．

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BAKERS SELHURST
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\section*{IWin．Baker，15w．Atalmart} | 3 or | 15 ohms． |
| ---: | ---: | ---: |
| c－p．5． 13,000 |  | c．p．s． Tin．Steren，Foam sue penalo，［2\％．，： 5 －1 1i，400


 1以in．De lise ． 7 gns
 1311．Bass 25w．20－14，1000

 pronfed paxolin coil formpr． 18 gas．

LOUDSPEAKERS P．M． 3 OHM．
 30／－：12in．R．，, $30 /-:$ EAH mas； STENTORIAN HF1012，I
रig．Hf゙く1．72／－： EXTENSION SPEAKER CABINETS 4 in ．， $15 / 6$ ；

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tubps with tailing emission instructions supplied．mains input． Inke A onitional soi and ign horosi


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 ath trimilirs，9／－；midist，z／6：with trimmers．9／－


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CONDENSERS．Anw Ntw－k，0．0f1 mid． 7 k








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Alumisium (ligaden, 18 s.w.g. Plain,
undrilled. 4 sides, rivetod corners,
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4/9; $9 \times 7 \mathrm{in},. 5 / 9: 11 \times 7 \mathrm{in}$. 6/9: $13 \times 9 \mathrm{in}$.
8/6: $14 \times 111 n . .10 / 6 ; 15 \times 141 n ., 12 / 6$.
Aluminium Panc•m, $18 \mathrm{~s} . \mathrm{w}, \mathrm{g} ., 12 \times 121 \mathrm{n}$.
4/6; $14 \times 91 \mathrm{n} ., 4 /-; 12 \times 8 \ln .3 / 6 ; 10 \times 7 \mathrm{in}$. .
2/3; $8 \times 6 i n . .2 /-; 6 \times 4 \ln . .1 / 6$.

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MAINS DROPPERS, Midget adjustable miders 0.3 A . 1,000 ohms, $6 /-; 0.2 \mathrm{~A}, 1,200$ ohtris, $6 / \cdot ;$ 0.15A, 1.500 ohma, $8 / \circ ; 0.1 \mathrm{~A}, \mathbf{2}, 000$ ohmes $8 / \mathrm{m}$. MIKE TRANSFORNERS, $50-1,3 / 8$
P.V.C. Covered Wire, single or gtranded, Rd, yd,
 $17 \times \rightarrow$ PER-FRET. Hold $10 /$ Haroon or siln, wite from $10 /=$ ft.: 2iin. wide 1 min 5 )- it.




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WINCLAIR "siIININNE" KAIIO Med. whve kit, 2 transistors, 2 diodes. earphone, ferrite acrial, Cabinot $3 \times 11 \times 11 \mathrm{n}, 49 / 6$, batt, $3 / 6$.

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# Practical Wireless 



Vol. XL No. 689 JULY, 1964


## The Role of Broadcasting

T1 HE recent events subsequent to the appearance of those electronic buccancers Radio Caroline and Radio Atlanta have raised many interesting points. At the time: of writing, the situation is still unresolved, although the fate of these floating "discophoriums " seems sealed by way of restrictive legislation. However, we do not intend to enter into the rights and wrongs of the affair. but to follow certain trains of thought.

Casting aside. therefore, any legal, moral or cultural considerations, it becomes manifestly clear that there exists a quite considerable public for the type of wares dispensed by Caroline and her sister craft. If there was ever any doubt, confirmation came by way of the extraordinary announcement from the BBC that their Saturday afternown sports broadcasts on the light Programme would be replaced by Caroline-type continuous pop music. Coming at the height of the controversy this was not only an admission of the drawing power of such broadcasts but even gave the impression of a panic measure.

Having established the existence of the demand for continuous pop music (though how any devotee can listef. to more than one station at once is baffing, for we already have Radio Luxembourg and others) let us go a stage further.

On a recent TV interview, a spokesman for Radio Caroline talked of a great and general requirement for continuous background music with the minimum of interrupion. His point (which is shared by others) was that TV is the medium for those wishing to sit and concentrate, whereas radio should be relegated more and more to providing only a non-stop supply of music which can be turned on and off like a tap.

So now we come to the fundamental issue: should this be the ultimate role of broadcasting?

The fact that pop music is cited is irrelevant; it could as well be any kind of music-indeed, so far as some are concerned, even perhaps traffic noises or sound effects! The need. apnarently, is to soak continually in sound like blotting paper.

Transistor radios, due to their advantage of portability, have unfortunately contributed something to this morbid craving for background music, or noise. It is common enough to hear them being played loudly while the "audience" virtually ignores its efforts to entertain. We sec the same thing in other places where somebody will feed a juke box with coins, only to become completely oblivious (perhaps wisely) to the less-thanahi-fi being pounded out, except in attempting to converse above the level of disorted watts. We also see it in the vogue for piped music at railway stations, in restaurants, clubs, pubs, lifts, etc. We shudder to think where it might be encountered next!

This pathological craving for noise for the sake of noise is a symptom of modern living and as such is the province of the psychiatrist, the neurologist and the sociologist. What we are concerned about is this tendency to regard sound broadcasting as a kind of super piped background music. Surely this is not the role of broadcasting?

Our next issue dated August will be published on July 7 th


## R.N. Radio Amateurs Donate Receiver

"WO serving members of the Royal Navy Amateur Radio Society
recently made a call at the home, in Crewe, of 18 -years-old

> NEWS AT HOME AND ABROAD

John Hall who was paralysed as a result of an accident while serving in the Royal Navy. The visit was made in response to an appeal for help afier another radio antatenr had informed the R.N.A.R.S. of John Hall's interest in wireles.s. and so on April 18th. G3SJQ and G3JFF (Sub-Lieutenant David Davies and Chief Radio Supervisor Michael Matthews respectively) arrived at his home with a radio receiver which local hams had volunteered to modify for his use.

With this receiver and the help of other radio amateurs in the area, John Hall is now well on the way to achieving his ambition of obtaining an amateur transmitting licence.

## EUROPE'S FIRST QUARTZ GROWING PLANT

THE first commercial quartz-growing plant in Europe has been established by Standard Telephones and Cables Limited at their recently opened quartz crystal factory at Harlow, Essex.

Using specially designed pressure vessels and control equipment, STC technicians have perfected a process which permits the growth in less than a month, of large pieces of radio quality quart $z$. suitable for use in the frequency control units used extensively in broadcast transmitters, telephone and communications equipnent. Against this the world's deposits of natural quartz took over $3,000,000$ years to form.

Whereas natural quartz stones are very irregular in shape and carry unwanted minerals and other inclusions, these man-made quartz crystals are geometrically shaped and virtually free of defects This new process also presents economical advantages, as natural quartz of suitable grade (which is obtained entirely from Brazil) is expensive and the large pieces necessary for manufacture into frequency control devices are rare.

The growing process of the cultured quartz takes place inside 10 ft . high, 12 in . diameter steel cylinders at pressures of about $24,0001 \mathrm{~b} . / \mathrm{sq}$. in. and at about $400^{\circ} \mathrm{C}$. Small, easily obtainable quartz chips are dissolved in caustic soda under these conditions and the quartz re-appears as a steady growth on special "seed" crystals placed in the. upper part of the cylinder. Crystals grow to the required size -about 11b.-in 21 to 24 days.


The quartz crystals shown in this illustration were grown in three weeks at STC's factory at Harlow.

## NEW TELEPHONE CABLE FOR S.E. ASIA

'THE first stage of the SouthEast Asia Commonwealth submarine telephone cable system (SEACOM) is to be laid in
the China Sea during the next six months. Three cableships will lay some 2.000 nautical miles of cable linking Hong Kong, Jessel-

## ADEN GETS NEW WIRELESS RECEIVING STATION

NEARING completion at Salt Pans in Aden is a wireless receiving station being built by Cable and Wireless (Mid East) Lid. as a sister station to the Company's existing transmitting station there. Between them. the two stations will provide a number of wireless services linking London with cities in various parts of the Middle East, Africa and India. These services will include public telegraph. leased circuits, telex and radio telephony. In addition, the pair will serve as a radio relay station on the London/Singapore, Hong Kong, Australia and Japan routes.

## Radio Safety Device for Swimmers

IINGINEERS in the U.S.S.R. have announced a "hydroacoustic" device which will automatically send out radio
distress si_nals should any mishap occur to a swimmer carrying it.

The device consists of a miniature transmitter-which can be instatled in a special pocket in a bathing cap-and a receiver on the shore. Should the swimmer stay under water longer than a minute. the transmitter will automatically send out distress signals, to be picked up on shore by the receiver, which will then give out light and sound alarms.

The transmitter itself consists of a sensor, a time relay, and a piezoceramic generator of ultrasonic waves and works with an accumulator.

## Safety Plan to Protect Consumers

() ${ }^{N}$ the eve of the Annual Conference (April 26th-29th) of the Radio and Television Retailer's Association and the Electrical Appliance Association. the President of the two Associa-" tions, Mr. A. R. Mitchell, explained a ncw "Safe Shopping" plan for the public. which was launched at this year's Conference.

The aim of the plan, as outlined by Mr. Mitchell. is to help protect the public from the increasing number of "apparently 'free" offers, switchiselling, excessive interest charges, 'vanishing salesmen' and other devices" appearing in the domestic electrical retail market.
Under the plan, members of the public shopping at EAA shops will receive a registration card with their purchases of major electrical appliances and on return of this card to the Association they will be enrolld on a new Consumer Register. The card will then be returned to them with the serial number under which they have been enrolled. They can then write to the Association with any enquiries they may have; for advice or to complain either about a product bought, or about service received, or about costs or either or both. Subsequently each enquiry will be investigated the app:opriate ones going before special committees uhich will then deal with them.

## New Branch of "Taking Book" Library

() N Friday, 1st May, Lord Fraser of Lonsdale, C.H., C.B.E., Vice-President of the Royal National Institute for the Blind and Chairman of St. Dunstan's, opened at Punch Street, Bolton, Lancashire, the Northern Branch of the Nuffield Talking Book Library for the Blind.
The new premises at Punch Street have been established to provide a localised library for the blind men and women of Scotland and the North of England because, as a result of recent technical advancements which have enabled the Library to accept a greatly increased membership, it has become irapossible for the original premises in Middlesex to cope for the whole country.

These technical advancements refer to the Library's tape machines which in 1959 replaced the existing stock of dise recordings of literary works, which was necessarily limited in eapacity. The present system, utilising a half-inch tape carrying no fewer than eighteen tracks, provides some twenty-one hours of reading time in one cassette.
The continued existence of the Talking Books Library-an entirely British innovation-is made possible to a large extent by the network of volunteers throughout the country who help to install and service Talking Book machines.

The rapid expansion of the Library membership now increases the need for these servicing volunteers and Mr. D. Finlay-Maxwell, Honorary Organiser of Servicing Volunteers. would be glad to hear from suitably qualified people within the Bolton Branch service area who would be prepared to help in this vital and rewarding work. Potential volunteers should write to him care of J. Gladstone and Co. Ltd., Wellington Mills, Huddersfied, Yorkshire.

# T R A N S I S T O R 25 Watt P.A. amplifier 

 BY A. J. SHORTTHERE are many occasions when an audio amplifier is required at places remote from an a.c. mains supply. To address crowds on sporting and gala occasions, for the modalation of mobile transmitters, and to control and inform the public at outdoor meetings, or for street advertising-these are but a few of the possible applications. A portable amplifier for use with electric guitars. now popular in dance bands, may be a further requirement.

It is, of course, practicable to operate a valve amplifier from a rotary or vibratory convertor. This system has the disadvantage, however, of being electrically inefficient, and large accumulators are necessary to provide the power for a relatively modest signal output. Methods of increasing the electrical efficiency of various systems, intro-duce-in their turn-further complications.

## Advantages of Transistors

A transistor amplifier offers a very attractive answer to the problem. Firstly, a high degree of electrical efficiency can be obtained. Secondly, with no warm-up time involved, the unit need only be switched on for the actual occasions when it is required. Thirdly. the unit is much smaller and lighter than an equivalent valve amplifier. Fourthly, with no fragile valves included, the unit can be made more robust and less liable to damage than a valve amplifier. Fifthly, transistors can be expected to have an appreciably longer life than valves when worked under comparable conditions.

## Design Requirements

It was decided that a useful level of power output for a simple. home constructed, transistor p.a. amplifier would be a nominal 25 W peak audio. This permits the use of a pair of a wide selection of power transistors now offered by the manufacturers. Costs involved at this Ievel are not excessive. and the remaining materials required can be readily obtaincd. If higher power levels are requited more power transistors will be needed to drive the outnut atage and the cost of the amplifier increase tharply.

Concerning frequency fange, it was decided that
a nominal $200 \mathrm{c} / \mathrm{s}$ to $7.000 \mathrm{c} / \mathrm{s}$, would be required. This has been achieved with reasonable linearity in the design. Some modest falling off of gain occurs at each end of this range, but the quality of reproduction under par. conditions is good.

The power supply requirements of an audio amplifier can be considerably reduced if the output stage is operated in Class B. as is now standard practice in transistor technique. A further advantage of using this mode of operation is that it cuts down the heat dissipation of the output transistors and reduces cooling problems.

In the prototype model of this amplifier it was considered unlikely that the amplifier would be actually producing its 25 W for more than a few moments at a time, and the cooling was reduced further to accommodite this factor. Should it be thought that the amplifier might be operated at maximum output for appreciable periods, the cooling of the power transistors should be improved accordingly. This can be achieved by mounting the transistors on heavy copper, rather than aluminium, increasing the area of contact between the copper and the casing of the amplifier. and increasing the thickness and size of the amplifier instrument case. As the amplifier case is being used as heat "sink" or radiator, care must be taken to see that it is always used in a well ventilated situation.

## The Circuit

The circuit consists of four stages: a.f. amplifier, phase splitter, push-pull driver, and output stage. For simnlicity and economy the phase splitter, and push-pull driver stages are directly coupled. The output stage drives the loudspeakers via the autotransformer.

The a.f. amplifier $\operatorname{Tr} 1$ is a high gain stage employing an NKT223 audio transistor in Class "A", working at about 1 mA . The circuit is comnletely orthodox and is fully temperature stable. The stage should be fed in a manner appropriate to the type of signal source employed.

In the circuit diagram shown, the appropriate components to suit the output from a crvstal nickup or crystal microphone are given. It will be found advisable fo keep the resistance of the volume control VR1 low to help to keen innut noises at a low level, particularly when the amplifier is used in a situation having low ambient background noise.

## The Second Stage

The second stage transistor $\operatorname{Tr} 2$ accepts the large output signal from the a.f. amplifier and, by means of a split load in collector and emitter circuits, produces two phase-opposed signals to be applied to the driver stage.

Of the available $8-9 \mathrm{~V}$ h.t. rail, two volts appear across cach of the two load resistors R 10 and R11, and 4 V across the transistor. This permits a maximum theoretical peak-to-peak signal of 4 V to be applied to the two halves of the following stage. The large bias, of just over 2V. applied to the base of this transistor is in accordance with the high emitter voltage.

As part of the load is in the emitter circuit of this stage, a large signal voltage excursion occurs
on the bane of Tris．The collector current and collector resistor of the first stage are ab selected as （1）make avalable this large volage swing（ $\pm 2 \mathrm{~V}$ ）．

## The Third Stage

The thira or driver stage consists of two transis－ tors Tr3．Tr4 working in Class A push－pull to provide drive for the output stage．These two transistors are directly coupled 10 emitter and collector of the preceding phase－splitter．and the steady voltages on these points provide the bias for the driver stage．

Decoupled emitter resistors adiust the standing current in these transistors to about 25 mA each． and the collectors are partially starved of h．t． voltage to keep the collector dissipation below the rated maximum for these components．The collec－ tors of these transistors are connected to separate primaries of the driver transformer T1，wound so that the d．c．flux of the collector currents tends to cancel out．The a．c．fields，however，heing in push－ pull，are additive and are apolied to the transfor－ mer secondary．The d．c．fux from the bias currents in the secondary is balanced out in the same way as in the primary．

The transistors in this stage should preferably be a matched pair．

## The Output Stage

The output stage consists of a pair of matched power transistors．Trs．Tro．in Class B push－pull． and these are counled to the loudsneakers via an auto－transformer T2．An auto－transformer is used here becatuse of the simplicity of construction of this component．The power transistors are cooled

## COMPONENTS LIST

## Resistors：

| R1 | M | R1I | 1.5 kO |
| :---: | :---: | :---: | :---: |
| R2 | 33kS」 | R12 | 470』．${ }^{\frac{1}{2} \mathrm{~W}}$ |
| R3 | 4.7 k S | RI3 | 10s |
| R4 | $4.7 \mathrm{k} \Omega$ | RI4 | 1005 2 |
| R5 | 1 k ， | R15 | $220 \Omega$ |
| R6 | 39k | R16 | $1 \cdot 5 \mathrm{k} \Omega$ |
| R7 | 100 kS | R17 | 20S，$\frac{1}{2} \mathrm{~W}$ |
| R8 | 18k』 | R18 | 100S |
| R9 | IMQ | R19 | 820 S |
| R10 | 1．5kg | R20 | 820 2 |

All $\pm 10 \%, \frac{1}{4} W$ except where otherwise stated． VRI 50kS carbon potentiometer．

## Capacitors：

| Cl | 22 pF silver mica |
| :--- | :--- |
| C 2 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 3 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 4 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 5 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 6 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 7 | 47 pF silver mical $2 \%$ |
| C 8 | $1,000 \mu \mathrm{~F}$ electrolytic 12 V |
| C 9 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 10 | $100 \mu \mathrm{~F}$ electrolytic 12 V |
| C 11 | $100 \mu \mathrm{~F}$ electrolytic 12 V |

## Transformers：

TI Driver transformer（see text）
T2 Output transformer（see text）

## Transistors：

Trl NKT223，Tr2 NKT223，Tr3 NKT22B，Tr4 NKT228，Tr5 NKT404，Tr6 NKT404
Fig．1：The circuit of the amplifier．

by a thick $\frac{1}{8} \mathrm{in}$. aluminium plate on which they are mounted, which is, in turn, bolted directly to the metal case holding the amplifier. The case then serves to cool the transistors. Mica washers and silicone grease between the cooling plate and the body of the transistor serve to provide electrical insulation while maintaining thermal conductivity. The mica washers and insulating bushes for the transistor mounting bolts are provided by the transistor manufacturers.

In the prototype it was found convenient to mount the output transformer a short distance from the amplifier. When this is done, care must be taken to ensure that the connecting leads are thick and of low resistance, as currents of about 3A flow during peak signals.

When the output transformer is mounted in the same case as the amplifier the cores of the driver and output transformers should be kept at right angles to one another.

## Construction

The first three stages are best constructed on a "home-made" printed circuit board, the full size layout of which is given in Fig. 2.

The copper coated laminate may


Fig. 2 (above): This illustration of the printed circuit is drawn full-scale and may be used as a template.
Fig. 3 (left): The layout of components.
be obtained from one of a number of advertisers in this magazine, and is cut to size with a fine hacksaw blade. The copper should then be brought to a bright shining tinish by scouring with "Vim" or other detergent scourer, and well dried.

A stencil of the printed circuit diagram should be made by cutting out and sticking Fig. 2, or a tracing thereof, to an ordinary postcard and cutting out the black portions with a sharp bladed penknife. The extracted pieces should be kept on one side to later locate the drilling holes.
The circuit diagram is then transferred to the copper laminate by drawing with a pencil through the cut-out portions of the stencil onto the copper. Once this has been done and checked, the printed circuit diagram on the copper can be filled in with a dark varnish stain, or dark french polish applied with an artist's brush. The identifying letters may also be marked in varnish if desired.

The varnish will now protect the covered copper from the etching solution to be used. The unprotected portions are to be dissolved away, so care must be taken to avoid greasy finger marks or varnish splashes on these parts.

## The Etching Process

The etching solution consists of $40 z$. Perric chloride, $10 z$ hydrochloric acid. 60 z . water, this solution can be made up by the local chemist.

A sufficient quantity of this solution should be poured in to a non-metallic container and the laminate floated, face down, on its surface. The rate of etching varies largely with temparature and may take 15 to 30 minutes or more. As soon as all trace of the unwanted copper has disappeared, the board should be well washed in running water to remove all traces of the etching solution.

When dry the board is drilled at the points indicated on the printed circuit diagram Fig. 2 and the varnish carefully scraped away with a sharp knife from the points to be soldered.

The components should then be inserted in position, as shown in Fig. 3 and soldered to the copper laminate. All loose ends should then be clipped off,


The driver transformer ready to be clamped into position.


Fig. 4: Winding details for the driver tronsformer, TI.
connecting leads inserted and soldered, and the completed printed circuit board mounted in the instrument case by nuts and bolts through the unused ends of the printed circuit board.

Ensure that the underside of the board is well clear of adjacent metal work. The board is now best removed from the case until all assembly and testing has been completed.

## The Driver Transformer

The driver transformer $T 1$ is quite simple and may be constructed in a few hours. It consists of two primaries of 480 turns each of 28 s.w.g. enamelled copper wire, and a centre-tapped secondary of $160+160$ turns of 26 s.w.g. enamelled copper wire, wound on a type HWR 10/16/5 English Electric "C" core pair, or equivalent, see Fig. 4.

The required dimensions of the core are: crosssectional area through centre limb, approximately 0.3 square inch. length of flux path approximately 5 in.. window area approx $\frac{3}{4}$ square inch as indicated in Fig. 5. A suitable core may be conveniently recovered from an old a.f. or output transformer. The laminations should be thin for minimum losses.


Fig. 5: Critical measurements for the transformer cores.
A cardboard bobbin should be first constructed to fit the core. This should be well dried and shellacked. The two primaries. of 480 turns each, are first wound in even layers, a layer of thin paper between each layer. Two layers of Sellotape should separate the two primaries from one another, and another two layers separate the primary from the secondary.

The secondary of $160+160$ turns is then wound, bringing out a connection to the centre tap, and the final assembly firmly overtaped with Sellotape. All windings are in the same direction and the ends of the primaries brought out through
the sides of the bobbin on the opposite side to those of the secondary.

The core should now be inserted. If " $C$ " cores are used the two halves must be clamped firmly together. If laminations are used they must be inserted alternately from opposite sides to avoid having an air gap in the core, the final laminations being wedged, if necessary, to avoid buzz and chatter when the transformer is used.

A small tag-board should now be strapped to the transformer to take the connections to the windings, and each winding identified on the tag board.

## The Output Transformer

For quality of reproduction, efficiency, and ease of construction, it was decided to design the output auto transformer T2 to generous proportions. The resulting performance, on test, was therefore found to be extremely satisfactory and the amount of wire winding involved was surprisingly small.

The prototype was required, primarily, to drive two $15 \Omega$ loudspeakers in parallel, but by means of a little juggling in design a most effective compromise to permit the use of alternatively one or two $15 \Omega$, two or eight $3!2$ loudspeakers, has been produced. The transformer consists of a single winding of 272 turns, tapped at 68 turns and 136 turns from one end as shown in Fig. 6. The two halves of the winding are wound with different sizes of wire to take into account the currents flowing.

The core consists of two pairs of English Electric HWR 50/32/5 "C" cores, or equivalent. The approximate dimensions of a suitable core are: Cross sectional area through centre limb. 1.8 square inches, length of flux path. 8 in . window area $2 \frac{1}{2}$ square inches, all approximate (refer to Fig. 5).

A suitable core of these dimensions, or the actual "C" cores themselves, can often be recovered from an ex-Government transformer, bought for a few shillings.

A suitable bobbin should be made to fit the core and well dried and shellacked. The end of the 18s.w.g. wire is passed through a hole made in the side of the former and 136 turns wound on in even layers, a thin layer of tissue paper between each layer. The end is brought out through another hole in the side of the former where it is connected to the end of the 15 s .w.g. wire.

Sixty-eight turns of 15 s.w.g. wire are now wound. and at the end of this portion a loop of wire, to provide the second tapping point. is passed through a suitable hole in the side of the former.

The final 68 turns are now wound and the whole assembly firmly taped to prevent buzz and chatter when the transformer is used. All windings should be wound in the same direction.

The core may now be placed in position and firmly clamped tooether to prevent gaos in the magnetic circuit. The wires from the winding can now have their ends cleaned and soldered firmly to a suitable tag strip.

## Alternative Cores

Both transformers have been designed with ample margins of tolerance to accept modest devia-
tions from the size of core specified, without change of number of turns. It should be appreciated, however, that in general, decrease in the size of the core requires an increase in the number of turns of all windings, and that finer gauges of wires will then be needed due to the smaller window areas. This, in turn, leads to increases in winding resistance, an undesirable characteristic.

The cores specified may be obtained from English Electric Co., Transformer Sales Dept, East Lancashire Road, Liverpool 10. Lancs.

(d) Eight $3 \Omega$ Loudspeakers

Fig. 6: Alternative output tronsformer arrangements.

## Assembly of the Components

Flexible flying leads are attached to points $C$, D, G, V, W and X. A red battery lead is attached to point $B$ and a black battery lead to point A, a flexible flying lead is also attached to each of these two points.

Connect the centre conductor of a length of microphone cable to point $H$ and the screening to point J. This cable goes to the volume control.
-continued on page 239

# model control <br> -------------------------receivers 

TWO RECEIVER DESIGNS FOR 27Mc/s BAND OPERATION BY F. G. RAYER

IWHE two model control receivers described here are for use with any home constructed or ready-made c.W. mordel control transmitter, working in the $27 \mathrm{Mc} / \mathrm{s}$ band. The actual transmitter may have one or two valves. the higher power obtained from two valves being needed when maximum range is required.

The first receiver to be described is of very simple type, and is easily adjusted. It is suitable for short range working only-indoors, in a garden, or when sailing a boat on a small pond.

Much greater range is obtainable with the super-


Fig. 1: A simple two-transistor receiver for short range work.


Fig. 2: The method of construction for the tronsistor receiver recommended by the author.
regenerative one-valve receiver. so it is suitable for planes and boats. It will also give good range with a low-power (hand held) transmitter.

## Transistor Receiver

The circuit of this simple two transistor receiver is shown in Fig. 1.

When the carrier signal is received, the diode causes the base of transistor Trl to become negative. Tril conducts, increasing the voltage drop in the collector circuit resistors (R1, VR1) so the base voltage of Tr 2 is more positive, releasing the relay RLA so that the relay contacts close.

The receiver parts can be assembled on a small piece of paxolin, or on a tag board (see Fig. 2). The layout is not important, though connections in the tuned circuit should be reasonably short.

The tuning coil can be self-supporting, and can have 12 turns of 18 s.w.g. bare or enameded wire. It is about $\frac{1}{2}$ in. inside diameter, and about $\frac{7}{8}$ in. long. The tapping 2 is four turns from end 3 , and the coll is supposted by soldering it to tags.
for best results. a good quality diode for $27 \mathrm{Mc} / \mathrm{s}$ or higher frequency is required. Most types. except cheap sub-standard surplus, are satisfactory, Trl is a high-gain audio or r.f. transistor, such as OC71. etc. Tir is a small output type transistor. such as OC72. The transistors operate as direct current amplifiers. their cut-off frequency is thus of no importance. but they should have high gain.

The relay is the usual model control type, with 1.000 to $5.000 \Omega 2$ windings. A $5 \mathrm{~mA}, 10 \mathrm{~mA}$ or similar meter, or multi-range meter. is placed in one battery lead for tuning purposes.

## Tuning Up

A vertical aerial. some 16 in . to 24 in . long, is connected to 1 on the coil. The transmitter aerial may be removed. With the transmitter and receiver near together, the 30 pF trimmer TC 1 is rotated until a dip in current is observed each time the transmitter is keyed. Tuning is quite flat, The transmitter aerial can then be added, and the receiver is moved a few yards away, and the $100 \mathrm{k} \Omega$ potentiometer and trimmer are set for best results.

Range should be sufficient for the purposes mentioned when used in conjunction with a twovalve transmitter. For maximum possible range, the tapping 2 may be modified a turn at a time, to suit the diode and Tri. The aerial may also be lengthened, and the turns on the coil can be


Fig. 3: A super-regenerative circuit for better reception.
adjusted so that resonance is obtained with the 30 pF trimmer nearly completely unscrewed.

## Super-regenerative Receiver

The circuit of a more sensitive receiver employing a single valve is shown in Fig. 3 .
The small coil L1 is tuned to $27 \mathrm{Mc} / \mathrm{s}$ by

adjusting its core and the 30 pF trimmer TCI. Regeneration is obtained from the feedback winding which is coupled to L1.

L2 is the quench coil. with its own feedback winding, and it tahes the valve in and out of oscillation.

Ihis super-regeneration receiver gives great sensitivity, and a good current change (to operate the relay even with a weak input signal.

The $27 \mathrm{Mc} / \mathrm{s}$ coil is wound with 26 s.w.g. enamelled wire, with turns side by side, as in Fig. 4. A space of about $\frac{1}{16}$ in. is left between 11 turn and 15 turn windings. Two cores are necessary in the coil. one inserted from each end. Each core is initially screwed in for about two-thirds of its length. Both windings are in the same direction, and touches of adhesive will help hold the ends.

A piece of wood $\frac{1}{2} \mathrm{in}$. in diameter and about $1 \frac{1}{4} \mathrm{in}$. long is used for the quench coil. Two paxolin discs. about $1 \frac{1}{2}$ in. in diameter, and with $\frac{1}{2}$ in. holes, are cemented on the wooden former, leaving 9 mm winding space, as in Fig. 5. Both windings are of 36 s.w.g. enamelled or silk covered wire, and are in the same direction.

The 600 turn winding begins at Q1, and ends at Q2. A strip of paper is then placed over this winding, and the 300 turn winding is put on, beginning at Q3 and terminating at Q4.


The receiver is built on a tag board or paxolin panel. refer to the layout shown in Fig. 6. Leads to the $27 \mathrm{Mc} / \mathrm{s}$ coil. 30 pF trimmer, valveholder, and 100 pF and 200 pF capacitors must be very short and direct. The quench coil can be held with a woodscrew.
Valveholder tag numbers in Fig. 3 are for a 3V4. Other valves may be experimented with. The best current change is obtained with a valve having a high mutual conductance.
The relay is the ordinary high resistance model control type, and a 10 mA or similar meter is added in the h.t. lead. Any $1 \frac{1}{2} \mathrm{~V}$ dry battery is satisfactory for the filament. A $67 \frac{1}{2} \mathrm{~V}$ miniature battery is necessary for h.t.
An aerial some 12 in . to 18 in . or so in length. is connected, and the receiver is placed near the
transmitter, the transmitter aerial being removed. The $30 p \mathrm{~F}$ trimmer in turned with an insulated tool. until the meter show's a current drop each time the transmitter key is closed. If this drop is extremely small. screw in core A a littie, and unserew the trimmer. to restore tuning. If no current drop can be found. serew in core $B$ a little. and again adjust the trimmer.

If the carrent drops when the transmitter key is closed, but does not rise immediately the hey as released, core $B$ may be ton far in. The inductance/ capacity ratio of circuit L1. TC1. depends on the positions of core $A$ and the 30 pF trimmer. while the degree of regeneration also depends on the extent to which core $B$ is serewed in. Adjusiments to obtain a good current drop are very easy


The prototype of the two-tronsistor receiver.

## The "Junior" Crystal + 2

## FIJRTIIER INHORMATION CONCERNING TRANSFORMERS

TWHE althor of the Junior Set +2 (P.W.. Mav 19(4), has asked us to point out that Messrs. Radiospares have changed the size and design of their T/T3 transformer. The new version is more efficient and will of course work perfectly in the receiver, but unfortunately will not lit in the space allocated for it on the board. If a larger sandwich box is used to house the receiver, there will then be no problem.

Should the new T/T3 transformer be used then refer to Fig. 12 and call either of the widely spaced tags red, and the other blue. Or the closely spaced tags take any outside one and call this yellow, and the next tag to it brown. The other two tags are not used. Red and blue may be changed round for best results.

The following alternative transformers are all absolutely suitable, will fit in the space provided, but need changed colour coding:

D189 ARDENTE, T22 (make unknown) (Messrs. Henries Radio Lid.).

LT/44. Eagle Products (Messrs. Sexson, 162 Grays $\ln$ R Road, W.C.1.).


The more sensitive one-volve receiver completed.
near the transmitter, but become progressively more critical, as the range is increased.

The aerial length also has considerable effect on regencration. so is best left unchanged, once satisfactory adjustments have been madr. Any change to the aerial, or core positions. will make re-tuning necessary. When adjustments have allowed a good current change to be oblained at some yards range, with no arial on the transmitter. the transmitter aerial can be added, and transmitter tuming checked.

If no possible adjustments give any current drop. connections to orle winding of the $27 \mathrm{Mc} / \mathrm{s}$ cont may have heen reversed. If L I is readily tunable to the transmitter signal, and only a very small current drop can be obtained, then comnections: one winting of the quench coil are probablv reversed.

| Coding (hanges <br> $($ Fig. 12) | LT/44 | Ardente <br> 184 | T22 |
| :---: | :---: | :---: | :---: |
| For Blue read | Green | Blue | Outside <br> Red |
| For Brown read | White | Green | Outside <br> Cireen |
| For Red read | Red | Yellow | Outside <br> Red |
| For Yellow read | Red <br> (lead near <br> White) | Red | Outside <br> Green |
| Fixing | as text <br> exactly | adhesive | adhesive |

The author would recommend LT/44 since this is almost identical in appearance to the $\mathrm{T} / \mathrm{T} 3$ originally specified.

We apologise for a small draughtsman's error in Fig. 12 where "D" on the coil should go to the junction R1/R2/C4 and not as shown to the battery positive line.

Beginners who wish to have help over any difficulties mav of course use our query service, enclosing the usual query couton.


These apart, there must still remain a host of listeners with (dare I say it) more serious tastes who confine their listening to the Home, Light and Third but who have yet to be converted to f.m. What could suit their requirements better than a preset v.h.f. receiver with a decent audio section and reasonably sized loudspeaker?

## All Tastes Catered for

Coming back to these transistor portables, it does seem that with circuit design now very much standardised, the presentation of the package has assumed greater importance than the contents. Styling is paramount. The feminine customer is wooed with delightful confections in exotic colour schemes plus all the trimmings: cherry red with gold grille; flame and white with polished aluminium embellishments: lovat green with silver trimthese are a few examples. The cases are sometimes padded and. perhaps, reach the ultimate in " a moulded case studded with nearly 200 sparkling imitation jewels". For her mate, the ouldoor sporting mood is exploited: less of the metallic trim, a predominance of leathercloth or. in the luxury class, an outer covering of real leather-saddle stitched of course.

I was amused by a technical achievement that sone current models feature, namely slow motion tuning. This is certainly a boon for getting those tricky stations right orn the nose; I think it inight become very popular!

## From a Surrey Wood

A reminder of yesteryear from an unexpected quarter. A few weeks ago the newly launched BBC-2 TV programme concluded its transmission one evening, somewhat surprisingly. by playing a recording of a nightingale singing in a wood as made during a live sound broadcast way back.

For the older viewers, this must have brought vividly to mind again those outside broadeasts of 30 -odd years ago. Perhaps our younger contemporaries will smile indulgently as I hark back to those warm summer evenings when with hushed expectancy we would listen as the announcer broke into the programme to say that we were about to be taken over to a Surrey wood, to hear the nightingale. Coaxed by a suitable musical accompaniment played by a lady cellist concealed in the wood. the bird would as often as not give a performance. On these occasions the Corporation and the listening nation submitted to the caprice of this little unseen songster,

Very tame compared with a view of tomorrow's dawn at Tokyo brought via satellite-oh yes certainly, but a salutary reminder of the progress made in a mere three decades!.


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# An Audio Mixer 

# A SIMPLE TRANSISTORISED UNIT 

BY G. R. ADDIS

T1HE following article describes the development and construction of an audio mixer employing the latest transistor circuit techniques. The circuit is such that it can be arranged to suit most needs. whether amateur or semi-professional.

Since it is likely that the equipment will be used for recording such items as live concerts or lectures, it will be appreciated that light weight, good performance and the utmost in reliability are very important factors governing the design of the unit. In order to keep the weight of the equipment to a minimum, it was decided that transistors rather than valves, should be bsed, hence diapensing with the need for large and heavy power supplies. Another desirable feature would be the facility to extend the number of channels as and when the


Fig. 1: A tronsistor in the emitter-follower configuration.
need arose with the minimum expense and trouble. In the final design these requirements have been fulfilled and the mixer meets the requirements admirably.

Preliminary experiments were carried out on a circuit based upon the enitter-follower confoguration. lig. 1. These proved to be unsuccessful, as expected. due to inter-channel modulation.

The first circuit being unsatisfactory. a circuit which is the combination of an earthed emitter stage and an emitter-follower was tried. A circuit


Fig. 2: The basic arrangement for a "Bow's brick" circuit.
such as this is known as a Bow's brick. The basic circuit is shown in Fig. ?

In this configuration, the brick requires both positive and negative power supplies with respect to earth. A more convenient arrangement is shown in Fig. 3, this requiring only a simple power supply such as a battery.

A circuit such as this gives a very stable voltage gain with negligible distortion. The voltage gain can be calculated from the formula:

$$
\text { Vgain }=\frac{R f b}{\operatorname{Rin}}
$$

thus with a feedback resistance of say $22 \mathrm{k} \Omega$, and an input resistance of say $2.2 \mathrm{k} \Omega$. a voltage gain of 10 would result, i.e. a gain of 20 dB . Conversely, if these 1 wo resistances were changed around, then there would be a voltage lose of 20dB.
the resistor $R$ is lor adfosting the bias conditions. since most of the eurrent flowing through the feedback resistor also How through resistor $R$. bhu cellsing the output to he negative with respect to earth and so providing 'Irl with a negative voltage on its emitter.

The transistors chesen for use in the mixer were the type (0) 45 . Thin was becalue of their avalat bility. good frequency response and liaily quiet operation. It was decided that the mixer should operate from a 9 V power supply. A suitable value of current flowing through Trl is in the order of


Fig. 3: An improvement on Fig. 2 requiring only a simple power supply.
0.5 mA . and that through Tr2 being about 1 mA . A suitable voltage on the emitter of $\mathrm{Tr}_{2} 2$ is 2 V positive with respect to earth, thus the voltage drop across R 6 will be 7 V .

$$
\text { Hence } R 6=\frac{7}{1 \times 10-} \Omega
$$

$$
\mathrm{R} 6 \bumpeq 6.8 \mathrm{k} \Omega
$$

Since the voltage on the emitter of $\operatorname{Tr} 2$ is 2 V , then the voltage on the collector of Tr 1 will be 2 V , thus if a current of 0.5 mA is flowing:
same voltage output when its individual "fade resistor" was at minimum value. Likewise, another variable resistance could be placed in series with all the inputs so as to provide a master volume control.

## Construction

Due to the relative simplicity of the mixer, the complete electronics, apart from the various input resistances, may be built on to a miniature 10 -way tagboard. The diagrammatical layout is shown in Fig. 5. Since it is likely that the reader may have

$$
\begin{aligned}
& \mathrm{R} 2=\frac{2}{0.5 \times 10-3} \Omega \\
& \mathrm{R} 2 \simeq 3.9 \mathrm{k} \Omega .
\end{aligned}
$$

Allowing for a potential dificrence of 3 V to be across Trl. then the voltage on the emitter of thes transistor will be $5 V$ with resp-ct to earth. .e. the voltage drop across R4 will be $4 V$.

$$
\text { hence } R 4=\frac{4}{0.5 \times 10-2} \Omega
$$

$$
R 4 \bumpeq 8.2 \mathrm{k} \Omega .
$$



Fig. 4: The circuit finally arrived at for the mixer.

The value of the feedback resistor, Rfb , was chosen to be $1 \mathrm{k} \Omega$, and since the potential difference between the emitter of Tr 2 and the base of Tr 1 is in the order of 3 V , then the current flowing through Rfb will be 3 mA . This current flows from the positive rail via R 3 ,

$$
\text { hence } R 3=\frac{4}{3 \times 10-^{3}} \Omega
$$

$$
\mathrm{R} 3 \bumpeq 1.2 \mathrm{k} \Omega .
$$

Since $R f b$ is $1 \mathbf{k} \Omega$, then for unity gain the value of the input resistance must also be $1 \mathrm{k} \Omega$, i.e. K in $=1 \mathrm{k}$ 几.

A capacitor is placed in series with the input resistance in order to prevent d.c. voltage getting on to the base of Tri from the input source. A suitable value for this component is $50 \mu \mathrm{~F}$, thus a good low-frequency response is to be expected. likewise, another capacitor, C3, is placed in the output circuit of Tr 2 , a suitable value for this component being $10 \mu \mathrm{~F}$. It is also necessary to decouple the emitter of $\operatorname{Tr} 1$ and an electrolytic capacitor having a value of $100 \mu \mathrm{~F}$ is used for this.

Since the gain of the brick is proportional to the input resistance, Rin, it is necessary to use a variable resistor with anti-logarithmic characteristics in series with the input in order to provide a logarithmic increase in volume as the control is turned in a clockwise direction. If a variable resistance having a value of $1 \mathrm{M} \Omega$ is used. then the output signal may be decreased by up to 60 dB from the original signal level; this being adequate for most purposes.

The final circuit of the mixer is given in Fig. 4. It will be appreciated that if variable resistances having a smaller value than the "fade resistances" were placed in series with each input, then the level of each channel could be adjusted to give the
his own ideas on the final arrangement of the input circuits, it is not proposed to give comprehensive details on the mechanical side of the construction.


Fig. 5: lllustrating the layout of companents when built on a 10 -way tag board.

## Performance

As stated earlier, the mixer is very reliable in operation. it has a good frequency response and a low signal-to-noise ratio. The measured figures below give a better indication of the performance. Frequency response . . . $25 \mathrm{c} / \mathrm{s}$ to above $200 \mathrm{kc} / \mathrm{s}$ $\pm 0.5 \mathrm{~dB}$.
Harmonic Distortion . . Not measurable.
Noise level . . . . . . . . .90dB below minimum sixnal output.


Fig. 1: The circuit of the unit.

# A TOP BAND CONVERTER 

THIS UNIT WILL ADAPT CERTAIN EX-GOVERNMENT RECEIVERS TO TUNE OVER THE $1.8-2 \mathrm{Mc} / \mathrm{s}$ AMATEUK BAND

BY J. CALLOW G3RBII

'HERE is on the market and in widespread use a large number of ex-government receivers which do not tune the $1 \cdot 8-2 \cdot 0 \mathrm{Me} / \mathrm{s}$ amateur band, most of them tuning only as low as 2.0M2/s. It was for this group of receivers that this converter was designed, the prototype being used on a BC. 454-B Command Receiver. The olltput of the converter is on $3.0 \mathrm{Mc} / \mathrm{s}$. and so it will be necessary for the receiver to tune this frequency.

## Circuit Description

Fig. 1 shows the circuit of the converter. The oscillator is tuned independently of the r.f. stages. thus eliminating troublesome alignment and the need for a three-gang tuning capacitor. VC3 tunes the oscillator over the range of $4 \cdot 8-5 \cdot 0 \mathrm{Mc} / \mathrm{s}$ and thus produces the required beat of $3.0 \mathrm{Mc} / \mathrm{s}$ throughout the band. The oscillator circuit is a Hartley, which requires no elaborate setting-up procedure. The valve used a 6C4 triode.

The two EF91 valves function as r.f. amplifier and mixer respectively, and the grid circuits are tuned by a twin-gang 500 pF capacitor (VC.1, 2). Osmor range 4 (trawler band) aerial and h.f. covils are ued. These coil, have iron cores and so no extermal trimmers are necessary. Ihe inductance
of the tuned windings is in the region of 35 uH . Selectivity in these circuits is of little importance. and it was only expense which prevented bandpals units being used.

The first EF91 is an orthodox r.f amplifier stage. having a $10 \mathrm{k} \Omega$ potentiometer VRI in its cathode to control the r.f. gain. For simplicity the screen of this valve is strapped to that of the mixer, and these screens receive their voltage through a single 22k! resistor. RI decoupled by capacitor C1.

The oscillater signal is applied to the grid of the mixer V2 via a 5 pF capacitor C4, The output signal is derived from a $3.0 \mathrm{Mc} / \mathrm{s}$ i.f. transformer (Fig. 2) which is wound in an old TV i.f. can.

No a.v.c. is applied to the converter. but if desired it can be applied by modifying the circuit

Fig. 2: Winding details of the $3 \cdot 0 \mathrm{Mc} / \mathrm{s}$ if. trans, ormer.
-



Fig. 3 (above): A modification to incorporate a.v.c.
Fig. 4 (right): The above-chassis layout of components.
as in Fig. 3, but it will be found necessary to connect the converter to the receiver with 4 -core cable, the extra conductor being connected to the a.v.c. linc of the receiver.

## Construction

The prototype was constructed on an openended chassis 6in. x 4 in . x 2 in . Fig. 4 shows the arrangement of components, though this is not critical, and need not be followed exactly as long as

care is taken to avoid any feedback. It must be noted however that the EFYl's have a high gain. and so only a little feedback is. necesnary to cause oscillation.

Owing to the superior selectivity of the converter, it was found necessary to use a slow motion


Fig. 5: The complete underchassis wiring diagram.
drive on the oscillator tuning capacitor. and even with this it was found easy to tune past a signal.

## Winding the I.F. Transformer

Details of the output i.f. transformer IFTI are shown in Fig. 2. All components for this were acquired from an old TV i.f. transformer. which was carefully cleaned of wax with methylated 9 pirits. The former was of $\frac{3}{3}$ in. diameter with iron cores. The windings consist of 72 turns of 28 s.w.g. enamelled copper wire, and are wound in two layers at each end of the former. Each winding is tuned with a 82 pF silver mica capacitor.

-continued from page 228 TRANSISTOR " 25 WATT P.A. AMPLIFIER"

The transistors are fitted to the printed circuit board and heat sink as shown in the following table:

| No. | Type | Base | Emitter | Collector |
| :---: | :---: | :---: | :---: | :---: |
| Tri | NKT223 | ToK | ToL | To M |
| Tr2 | NKT223 | To N | To P | To R |
| Tr3 | NKT228 | To R | To T | To Driver |
| Tr4 | NKT228 | To P | Tos | To Driver Transformer |
| Tr5 | NKT404 | Fitted to Cooling Fin |  |  |
| Tr6 | NKT404 | " | " . | , |

The printed circuit hoard and driver transformer can now be mounted in the case, carthing the case, il desired, to point $W$.

Connect the collectors of Tr3 and Tr4 to the start of one primary winding of the driver transformer and the finish of the other primary winding. The other end of the winding to which the collector of Tr3 is connected goes to the llying lead from point $D$. The other end of the

The tuned circuit in the oscillator consists, of a 30 pF tuning capacitor VC3 (such as obtained from an $\mathrm{RF}_{2} 7$ unit) shunted with a fixed capacitor of about 150 pF . C7, and an inductance L3. This inductance consists, of 17 turns of $24 \mathrm{~s} . \mathrm{w}$.g. enamelled copper wire close wound on a $\frac{1}{2}$ in. diameter former. tapped 5 turns from the earthy end. The inductance value is about $6 \mu \mathrm{H}$.

## Setting Up

The converter requires 6.3 V at 0.9 A and $200-300 \mathrm{~V}$ h.t. at about 30 mA . Such requirements can be furnished by the power pack on most receivers.

When the converter has been allowed to warm up for about 15 minutes, set the oscillator tuning capacitor VC3 to the half meshed position, tune the station receiver to $4.9 \mathrm{Mc} / \mathrm{s}$, and "zero" the oscillator in on this frequency by means of the oscillator padder TCl.

Then tune the receiver to $3.0 \mathrm{Mc} / \mathrm{s}$. connect an acrial to the converter, and couple the output of the converter to the receiver. The twin-gang tuning capacitor should be adjusted so that it is a little more than half meshed. The cores of the i.f. transformer are then peaked up. Next peak up the twin-gang tuning capacitor, and linally the cores in the acrial and h.f. coils. When all the cores have been properly adjusted they should be sealed with a drop of wax.
The 30 pF oscillator capacitor VC2 is the main tuning control of the converter, the twin-gang capacitor being used to peak up the sigrals. With a little practice and experience the converter will function well, and the selectivity of the combined equipment will be found far superior to that of any single conversion set.
winding to which the collector of Tr 4 is connected goes to the flying lead from point $C$.
The centre tap of the driver transformer secondary should now be connected to the fyying lead from point $V$, and the remaining two connections taken to the bases of the output transistors.

The flying lead from point $B$ should be taken to the emitters of both output transistors, and the collectors of the output transistors connected to the output transformer. together with the flying lead from point A . in accordance with the diagram Fig. 6. Connect the flying lead from X to the collector of Tr3. The flying lead from point $G$ should not be connected at this stage.
Check all wiring carefully and when satisfied that all is correct connect to a 12 V accumulator (observing nolarity) via a 1 A fuse.
The amnlifier should now work. and if satisfac1ory the fuse can be exchanged for one of 5 A rating and the feedback loov connection from point $G$ connected. To establish the correct point of connection test the flying lead from point $G$ on the base of each of the outnut transistors in turn. The correct connection will produce a reduction in amplifier sensitivity while the incorrect connection produces a vigorous howl. It will be observed that a mall internal feedback loon containine a $1 \mathrm{M} \Omega$ rescictor and a 47 nF caparitor connected to the base of Tr4 already exists. This is to balance out the internal collector to hase feedback of Tr4 and ensure stability of the $\operatorname{Tr} 2, \operatorname{Tr} 3 . \operatorname{Tr} 4$ network.


AS mentioned last month, several stations broadcast special programmes giving details of reception conditions and frequency changes. Here are details of some of them. All times are in GMT.
Radio Australia carries a programme called "Australian DX-ers Calling'". This deals exclusively with broadcast band stations and lasts ten minutes. It is beamed to Great Britain at 0715 on Sundays over $9,570 \mathrm{kc} / \mathrm{s}$ and $11,710 \mathrm{kc} / \mathrm{s}$. This programme is especiatly usefu! for keeping up to date with Asian and Far Easiern developments.

Ham band enthusiasts are catered for by Radio Sofia, Bulgaria, on Friday evenings at about 1445 and 2145. Best reception at hoth times is on $6.070 \mathrm{kc} / \mathrm{s}$. A similar type of programme is broadcast by Radio Prague. (zechoslovakiat. on the first Monday of each month at 1915 and 2215. The 1915 transmission comes in well on $7.345 \mathrm{kc} / \mathrm{s}$, whilst the 2215 transmission is on $1,286 \mathrm{kc} / \mathrm{s}$ medium wave.

One of the best programmes on the broadeast bands is that transmitted by Radio Denmark. Information is usually supplied by O. Lund Johansen, the World Radio/TV Handbook editor. Transmission times are Tuesday at 1920 on $15.165 \mathrm{kc} / \mathrm{s}$ and 2035 on $9,520 \mathrm{kc} / \mathrm{s}$, Wednesday 0220 on $9.520 \mathrm{kc} / \mathrm{s}$ and Thursday on $15.165 \mathrm{kc} / \mathrm{s}$ at 0950 and 1520. Reception, unfortunately, is often difficult as none of the transmissions are beamed to the British Isles or Europe.

Both ham and broadcast band listeners are catered for by Radio Nederland's programme "DX Juke Box". This is transmitted in all its English transmissions on Thursdays. Details of these transmissions are given in this month's schedule change news.
"Sweden Calling DX-ers" is one of the most comprehensive DX programmes. It is transmitted by Radio Sweden during the last 15 minutes of its English transmissions on Tuesdays. These are beamed to Europe from $2200-2230$ on $6,065 \mathrm{kc} / \mathrm{s}$ and $2330-2400$ on $1.178 \mathrm{kc} / \mathrm{s}$ medium wave. Some of the other beams can also be picked up in the U.K.

The second half of the Swiss Broadcasting Corporation"s English transmission on Saturdays is given over to a programme called "Swiss Short Wave Merry-go-Round ". On the first and third Saturdays each month there is a report on stations in the broadcast bands, whilst on the others amateur band topics are dealt with. The U.K. transmission is from 1845-1945 on 7,110/ $9,665 \mathrm{kc} / \mathrm{s}$.
The programmes mentioned so far are those which will probably be of most use to the average listener. There are also programmes from the Canadian Broadcasting Corporation. Radio Finland, Deutsche Welle, BBC, Radio Budapest,
N.H.K. (Japan), Radio Warsaw, Radio New York Worldwide and the Voice of America.

## Jamming

Good news for all broadcast band listeners comes in an announcement from the BBC on the subject of jamming. It says that all its East European transmissions, except those in Bulgarian, are now free of jamming. Let's hope this spreads and other stations transmissions go unblocked. The troable with jamming is that it is nonselective and often blocks out stations other than the one it is aimed at.

## Ham Band News

According to Radio Sofia the $21 \mathrm{me} / \mathrm{s}$ ham band is now good for working to the Far East. If you want to hear a station in Franz-Josef Land, Sofia suggests trying $14.040 \mathrm{kc} / \mathrm{s}$ or $14.060 \mathrm{kc} / \mathrm{s}$, where UAIKUD is working in CW. This station will USL through the Russian bureau, P.O.B.88, Moscow. A blot in the ham world is a report that the lebanon has suspended all licences.

VR4C N in the Solomon Islands will be working from 1200 on single sideband around $14.320 \mathrm{kc} / \mathrm{s}$ for about the next eight months according to the Swiss Broadcasting Corporation. They also report 20 m as the best DX band at present.

## B.C. Band News

In the broadcast bands more summer schedules have come to hand. Radio Nederland (P.O.B.222. Hilversum. Holland) is completely rearranging its schedule with the commencement of transmissions from its relay station in the Netherlands Antilles. New frequencies being used are $17.720 / 15.440 / 15$, $435 / 11,960 / 9.705 / 7,210 / 7,165 / 7,110 \mathrm{kc} / \mathrm{s}$. European English transmissions are from 20002050 on $6.020 \mathrm{kc} / \mathrm{s}$ repeated from $2100-2150$ in the 19 and 25 m bands. "Happy Station "programmes on Sundays have also been extended. A full schedule is available free of charge from this station.

Bernard Attrell. of Winchester, reports that the Windward Islands Broadcasting Service is back again to the U.K. on $15.085 \mathrm{kc} / \mathrm{s}$ from $2130-2230$. The signal is a very good one, especially when you remember the station is using only 5 kW . When reporting to this station there is no meed to send reply postage. Its address is St. George's, Grenada, W.I.

The Canadian Broadcasting Corporation, P.O.B. 6000. Montreat. is now using $5,970 \mathrm{kc} / \mathrm{s}$ and 4.625 kc , f for the whole of its English transmission to Australasia. This is on the air from 07150850.

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T10 many readers this article may seem to be going back a few years, and in a way this is true, for the first preliminary layout experiments of components was made by the author quite two years ago.

The small miniature components were purchased with the intention of constructing a portable of modest size which would he capabie of good reception on the three main BBC transmissions, using a frame aerial, and also to be capable of picking up short-wave transmissions by the inclusion of the appropriate short-wave coils and a tubular-type aerial.

These aims have been achieved and under good reception conditions quite a good selection of short-wave stations can be received on the two short-wave bands, whilst with a good aerial, wire twisted round the aerial tube, long-distance American transmissions have been received at, of course, the appropriate times.

The author must admit that the chassis presents a somewhat "dreadnought" appearance but the whole is the outcome of a series of additions which have taken place during the two years work on the set. 11 is, in fact. a prototype of an amateur constructor. The mechanical construction could no doubt be simplified in many respecis but not with regard to the general layout of component parts. which has received much thought, particularly in the designing for shorl and direct wiring. a necessity for the short wave bands.

So the author gives herewith drawing and constructive notes of this quite useful portable
in that it may tempt some readers to give these grand little valves a go again.

## CONSTRUCTIONAL DETAILS

As the author anticipated quite a lot of experimental work in connection with coils and frame aerials it was decided to matie the design such that the main tuning components were easily dismanted and replaced after the various modifications. For this reason a separate coil unit, complete with switching, was designed.

The several capacitors and resistors forming the antomatic volume control circuit were made in a unit form, as also the padding capacitors, these being neatly arranged in a "bank" 'mmediately adjoining the coil unit.
Again, for ease of removal the chassis was fitted with bottom flanges to engage into recesses in the frame aerial unit. With this arrangement and with but four soldered connections necessary, and with the two batteries in position. the set is completely self-eontained and workable. It can he slipped into the cabinet in a few seconds, however. by removing the tuning dial ard the three control knobs. By the way, an extension spindle with coupling is necessary for the main tuning capacitor.

## THE CIRCUIT

The theoretical circhit dagram is shown in fig. 1 and is a conventional woerhet arrangement of four valves: frequency changer, intermediate

| COMPONENTS LIST | Transformers: <br> TI L.F. transformer (Radiospares Midget 3-1) T2 Output transformer (Elstone Type MO/T) IFTI, 2 Miniature $465 \mathrm{kc} / \mathrm{s}$ i.f. transformer |
| :---: | :---: |
| Resistors: | (Repanco) |
| $R 1$ $33 \mathrm{k} \Omega$ $R 6$ $2 M \Omega$ <br> $R 2$ $5 \mathrm{k} \Omega$ $R 7$ $4.7 \mathrm{k} \Omega$ | Inductors: |
| $\begin{array}{lll}\text { R2 } & 5 \mathrm{k} \Omega 2 & \text { R7 } \\ \text { R3 } & 100 \mathrm{k} \Omega & 4.7 \mathrm{k} \Omega \\ \text { R } & \text { R8 } & 33 \mathrm{k} \text { ( }\end{array}$ | LI Long wave frame aerial 2 |
|  | L2 Medium wave frame aerial $f$ see text |
| $\begin{array}{llll}\text { R4 } & \text { IM } \Omega & \text { R9 } & 500 \Omega\end{array}$ | L3 Short wave 1 aerial coil (Osmor QA2) |
| R5 IMS2 ${ }^{\text {All }}+10 \% \cdot \frac{1}{4} \mathrm{~W}$ carbon | L4 Short wave 2 aerial coil (Osmor QAI) |
| VRI IMS carbon potentiometer, miniature | L5 Long wave oscillator coils (Osmor QO9) |
| VRI IMS carbon potentiometer, miniature button type | L6 Medium wave oscillator coils (Osmor QO8) |
| Capacitors: | L8 Short wave 1 oscillator coils (Osmor QO2) |
| Cl 180 pF silver mica | Switches: |
| C2 $0.01 \mu \mathrm{~F}$ paper | SI Single pole, 4-way wafer switch |
| C3 $0.1 \mu \mathrm{~F}$ sub-miniature type | 52 Three pole, 4 -way wafer switch with long |
| C4 0.01 $\mu \mathrm{F}$ miniature type | shaft |
| C5 150pF silver mica | S3 Double pole on,off toggle switch |
| C6 150pF silver mica | Valves: |
| C7 0.01 fF miniature type | VI DK92 V3 DAF91 |
| C8 0.01 14 F miniature type | V2 DF91 V4 NI/ |
| C9 $100 \mu \mathrm{~F}$ electrolytic 25 V | Miscellaneous: |
| C10 $8, \mu \mathrm{~F}$ electrolytic 150 V | Four BG7 valve holders with screening cans. |
| VCI 500 pF \} midget twin gang, with centre | Aluminium for chassis. etc. Perspex, $\frac{1}{8}$ in. thick |
| VC2 500 pF \} screen and long shaft | for coil unit. Tagboards. Dial $2 \frac{3}{4} \mathrm{in}$. dia. (push-on |
| VC3 25 pF \% $\}$ twin gang | type). Three knobs. Wire for coils (see text). |
| VC4 25 pF$\}^{\text {twin gang }}$ | Paxolin tubing $\frac{1}{2}$ in. dia. Plastic covered connec* |
| TCl. $4 \quad 500 \mathrm{pF}$ compression type trimmers | ting wire. Loudspeaker 2 in . dia. |

frequency amplifier, detector and audio frequency amplifier and pentode ounput stage.

The input from the long or medium wave frame aerials goes to the control gricl ( (i3) of V1, the frequency changer, whilst for the two short wave bands two coils are switched in and work in conjunction with a tubular acrial rod.

Four oscillator coils are used, the oscillator grid and oscillator anode windings of these being switched by S2h and S2c.

The intermediate frequency is the normal $465 \mathrm{kc} / \mathrm{s}$ and is fed to $\mathrm{V}_{2}$ via the miniature i.f. transformer $1 F T \mathrm{I}$. Here the $465 \mathrm{kc} / \mathrm{s}$ signal is amplified and the output at 1 FT2 is theal


Fig. 1: The circuit diagram of the Chelmer.
passed to the diode of V3 via the volume control VR1.

A miniature audio transformer T1. ratio $1: 3$, couples V3 to the power output pentode V4.

Automatic bias voltage is developed across R9. It will be noted that h.t. - and l.t. + are both switched off and on by the two-pole switch S3, thus preventing battery drain when the set is off.

It will be seen from the diagram that switch S1 is separate from $\mathbf{S}_{2}$.

Sl is used primarily for connecting the tubular aerial to the two short wave aerial coils. L3, L4, as it was found that the frame .aerials were adequate for the medium and long bands without any aerial. However, there is a slight gain, particularly on some of the Continental stations, by using the tubular aerial and, if desired, two wires
can go from switch $S 1$ to the grid ends of the frames, thus making a four-point change for the aerial.

It was not found desirable to use a.v.c. for the short wave coils, but if the reader wishes to include this the bottom end of the grid coils in the coil unit will have to be joined to the a.v.c. line instead of being earthed.

Above the main tuning (ganged) capacitor VCl, VC2, will be seen two smaller (ganged) vernier capacitors, VC3, VC4, for each of the two sections. This vernier will be found most useful for the short wave bands.

The author tried resistance coupling from V3 to V4. but transformer coupling was found much more satisfactory with the 90 V h.t. only available. A good-quality component is desirable both for T1


Fig. $2 a$ (above): Dimensions of the main chassis; $b$ (below left): drilling details of the

and for the matching transformer T2
Note particularly the filament connections to V4. The pentodes in this series of valves have a tapped filament for either 1.4 V or 2.8 V working. The circuit shows as connected for 1.4 V use.

## MAKING THE CHASSIS

The original chassis was formed on a standard aluminium box type chassis measuring 6 in . $x 2 \frac{1}{2} \mathrm{in}$. $x 2 \frac{1}{2} i n$. and to this stand-off legs and panels were added during the various stages of modification.

Many constructors will no doubt prefer to make up the chassis from a single sheet of aluminium and so full details are given in Fig. 2. The aluminium should be about $\frac{1}{16}$ in. thick.

It will be seen that in addition. to the main chassis piece there are two stand-otf panels (Nos. I and 2 in Fig. 2b) and an 1.f. screen.

The stand-off panels might possibly be dispensed with but the trouble of cramping below the chassis deck would have to be reckoned with if the on/otf switch and potentiometer were fitted inside.


The diagrams are fully dimensioned, although generally sizes have been given for centres. Diameters of holes should be checked with the actual components to be used. Sometimes makers provide templates and these should be used for marking-this applies particularly to i.f. transformers.

An "exploded" view of the chassis assembly is given in Fig. 2c.

The loudspeaker part is mounted on two lengths of aluminium curtain runner section bolted to the chassis, the speaker unit being secured to them by four lengths of 2B.A. screwed rods. The speaker can thus be adjusted to the required projection from the chassis. It will be seen that the twin vernier capacitor is mounted immediately below the main ganged capacitor, allowing very short connections and ease of operation. Both gangs require short extension spindles.

The aerial switch is just ahove the coil unit switch. It would be possible to combine the aerial and coil-change switches, using a four-pole, fourway switch in the coil unit. but such a switch will be found rather complicated, having tags on either
side of the wafer, whilst a ganged-type wafer switch takes up too much room for the coil unit.
The miniature loudspeaker is in the centre, mounted on its supports.

## THE COIL UNIT

Many readers may feel that this rather aweinspiring unit is quite beyond their scope. However, when once the framework of this unit is made up. with the wafer wavechange switch fitted, the wiring is progressive and, if a definite procedure is adhered to, it is not beyond the average


Fig. 3a (left): A general assembly view of the coil unit; $b$ (above): details of the end piece of the unit carrying tag board "c"; c (right): details of other end piece.
radio amateur. This procedure will be given later.

Full dimensions are given in Fig. 3 for the framework of the coil unit, including the screen.
Perspex sheet is a nice material to work with, although there is no reason why aluminium shect could not be substituted. apart from the problem of fixing the tag boards $A$ and $B$.

The general assembly is shown in Fig. 3a. Two lengths of 2B.A. brass serewed rod act as tie bars to the front and back plates and also form support for the aluminium screen. the lug pieces of the latter heing pressed tightly over the serewed rod with pliers. Thin sereen is. of course, situated at the back cnd. leaving a space at the front end for the switch and the wiring from it and from the coils.
When the two tag board, $A$ and $B$ have been
fitted the wiring-up of the coil unit can be commenced and it is wise to mark on the tag boards the connecting labels first. Thus "grid" (control), "OSCA", "OSC grid" and "h.t.". Also the tag board B should be labelled: "Aerial,"" Long". " Med..." "Short 1", "Short 2 ". The "Long " and "Mcd." are, of course, frame aerial connections in this instance.

First, by trial with a small battery and flashlamp, find the correct set of four contacts on the switch for each pole if this cannot be ascertained by visual means. Each set of tags should be colour spotted with four different colours, with the same colour for the pole tag itself. Thus we will have, say, red spots for OSC/anode connections, green for OSC/grid and white for control grid. This is a great help when wiring the coils. but care should still be exercised to get each coil correctly connected to the right tags for long, medium, short 1 and short 2 changeover.

Commence with wires from each of the main


A view of the completed coll unil.

Supports for speaker unit

kindly. Once securely soldered, however, the material is quite strong and a good insulator.

## THE MAIN ASSEMBLY

The back view given in Fig. 4 shows clearly the neat arrangement of components and these, being in the form of separate units. simplifies the below-chassis wiring. The units are also quickly removed, necessitating only a few minutes with the soldering iron.

A main h.t. positise bar has been fitted from the tag board on the left and is soldered to the appropriate h.t. leg of the first i.f. transformer. This bar, which has short lengths of sleeving threaded on, allows various h.t. positive, points available for connections. If red sleeving is used it will not be confused with the nearhy chassis wire (earth), also a useful arrangement.

This latter bar, together with its duplicate at the bottom of chassis, should, of course, be very thick gauge tinned eopper wire.: Certain wires, however, will be more conveniently connected components.
poles of the switch to control grid. "OSC anode" and "OSC grid" tags on tag board A, using fairly large gauge flex or stranded wire. A plastic insulated flex is less bulky than electric wiring type flex. Then continue by fixing the long wave coil oscillator, followed by the medium wave oscillator coil.

A good hot electric soldering iron of the pencil pattern is useful and it is advisable to wrap a small piece of wet or damp fabric round the coil tage as the plastic material used does not take heat
direct to chassis. Holes will be drilled for these and soldering tags bolted to the chassis. Such points are shown for V3 and V4 in the wiring view. Others could be so connected, if preferred, to common earth points.
On the right is seen the coil unit and from the small tag board on it go wires to the appropriate padder capacitors, the other ends of which are. to chassis.

CONTINUED NEXT MONTH

# MPROVED <br> CRCUIIS 

T1HE greatest enemies of the amateur radio and electronics enthusiast are (1) financial limitations, (2) lack of availability of components on the commercial market and (3) limitations in the operating characteristics of many of the components that are available.

Relays are typical components that can be considered as falling into all three of the above categories in that while some of the not very sensitive types may be purchased for only a few shillings, the price of those that will work from only a volt or two and have a resistance of thousands of ohms is quite prohibitive to many readers.

Situations often arise in which the experimenter requires a relay that can be operated from a current of only a few micro-amperes, but such components are just not available although they are, in fact. manufactured for the industry. Moving coil types are available on the surplus market, but these generally require a current of between one and two hundred micro-amperes to make them operate-which is too heavy a current to satisfy the above requirement.

## Critical On/Off Ratio

One of the most severe limitations in the characteristics of relays is the wide difference in
where a phototransistor or cadnium sulphide photocell is used to turn on a light when the daylight falls below a certain intensity and is intended to turn the lamp off again when the daylight rises back to that intensity, it is often found that due to the large difference in the relay on/off voltages the lamp often stays on almost permanently.
in this article it is intended to show simple circuits which are designed to overcome to a large degree these three enemies of the enthusiast when applied to relays.

## Increasing the Sensitivity

Transistors may be used to increase the effective sensitivity of both moving coil instruments and ordinary relays. Fig. 1 shows three circuits that may be used to this end. It can be seen that no biasing networks are included in the circuits and these have been found to be quite unnecessary in this particular application

In the circuits used as prototypes a relay with a coil resistance of 670 as used which required a current of 14 mA to "drop in", and sensitivity figures quoted will refer to this relay. It must, however, be pointed out that any relay that has characteristics falling within the practical voltage and



Fig. 1: Increasing the sensitivity of a relay with the aid of transistors.
the pull-on and the drop-out voltages. For example, it is found that a typical relay may pull on at $8 \frac{1}{2} \mathrm{~V}$ and drop out again only when the voltage is reduced to $1 \frac{1}{2} \mathrm{~V}$, an on/off ratio of nearly 6:1.

In many cases this on/off ratio, which gencrally lies between $3: 1$ and $4: 1$, may prove to be too high to give satisfactory operation of a piece of equipment. In a light-operated unit, for example,
current ranges of the transistor with which it is used will be found to work satisfactorily using the circuits shown. Red Spots (a very cheap and excellent general purpose transistor) were used, but many other transistors are suitable.

The relay is shown connected in the emitter circuft of the transistors, but it can equally well be used in the collector circuit with no noticeable change in performance.

# for Constructor 

Referring to Fig. 1a, the circuit must be connected up as shown and the triggering voltage applicd between the base and emitter. It is important to notice this point as if the trigger voltage is conneeted between the base and the positive supply line a very much greater voltage will be needed to cause the relay to operate.

In the test circuit it was found that the relav operated with a trigger voltage of 310 mV and a current of $2.6 \mathrm{~m} A$. The drop-out voltage was 250 mV which represents a very low an /oll ratio.

The cireuit in Fig. 1b. using iwo transistors, was found to operate from a voltage of 150 mV at $20 \mu \mathrm{~A}$ with the trigger voltage applied between the
prevent relay chatter. the component values being selected to suit the frequency of operation.

The second point to note is that if a number of relay circuits of this kind are to he used in an jnstrument, then each one must have its own separate power supply-and in any case no relay unit must have a supply common with the rest of the instrument circuitry.

If a barge number of such units are to be used in one instrument it would be as well to build a rectifier and smoothing network into each unit and drive cach one fron its own secondary of a transformer or inverter unit.

The third point is that all transisters used must



(c)

Fig. 2 (o): An arrangement to suit either positive or negative input signals; (b): to suit a.c. inputs.
(c): Diodes arranged as trigger voltoge limiters.
emitter and base of $\operatorname{Tr} 2$, while the circuit using three transistors shown in Fig. Ic operates from a mere 50 mV at $10 \mu \mathrm{~A}$.

All voltage rcadings were taken using a 20,0002 per volt meter on its 2.5 V ra.gge. It can be seen that at these low operating voltages and current the input to the transistors presents quite a high impedance.

## Triggering Polarity

Three points must be noted in particular about these circuits. First, they will only operate from one polarity of trigger signal, the hase forming the negative connection and the emitter the positive.

If it is required that the circuit operates from either positive or negative signals, a briuge rectificr must be connected across the input, as shown in Fig, 2a. Alternatively if the circuit is required to operate from a.c., a bridge rectifier and smoothing circuit must be used across the inputs as shown in Fig. 2b. The smoothing circuit is necessary to
have very high resistance between emitter and collector, otherwise a small current in one transistor may be sufficient to cause a sufficiently high current to flow through succceding transistors to prevent the relay from dropping out once it has been operated. If a "leaky" transistor must be used then it should be used in the Trl position, and then only if the leakage current is well below the rclay drop-out current.

## Checking the Transistors

It would be as well to check all transistors before connecting them in circuit in the usual way.

This can be performed by using an ohmmeter or a battery. a limiting resistor and a sensitive current meter in scries. Check that (1) a high resistance exists between the emitter and collector with the check leads in hoth polarities: (2) that with one lead connected to the base a high resistance exists between it and both the emitter and the collector: (3) that when the other lead is connected to the
base giving a reversal of polarity a low resistance exists between it and both the emitter and the collector.

The voltage ratings of the transistor must not, of course. be exceeded during these tests. A not very " with it " friend of mine once tested half-a-dozen transistors in this way with a 250 V Megger and was quite disgusted when none of them passed the test.

## Safety Diodes

A simple precaution to prevent trigger voltages of excessive amplitude from causing excessive current to flow in the transistors, with subsequent damage, is shown in Fig. 2c.

Here two germanium diodes are connected in parallel, anode to cathode and shunted across .he input. When only a few millivolts are applied to the diodes they act as a very high resistance and have almost no shunting effect. but as the voltage is increased their resistance falls, with a subsequent increase in shunting effect until when about half a volt is applied they act almost as a short circuit.
If the trigger voltage is to be applied with one polarity only, with no danger of a reversal of polarity, only one diode need be connected across the input as a precautionary measure.
It is often required that a time delay be incorporated so that either the relays will not click in until a certain time has elapsed from the moment of the trigger voltage being applied, or alternatively, they will click in as soon as the voltage is applied but drop out again after a certain time 'las elapsed.
With the more sensitive relays it is simply a matter of connecting a resistor and a capacitor of the

Fig. 3: Roising the input impedance still further by the use of two tronsistors; (a): 0 "delay-before-operate" circuit; (b): on "on-deloy-off" arrangement.
correct value and in the correct configuration directly to the relay in order to achieve this, but such relays are generally rather exsensive and it is with. methods of saving the amateur money that this article is primarily concerned.

Very cheap relays that require a comparatively high operating current may be used in conjunction with a couple of Red Spot or similar transistors to achieve the same or even better results at a fraction of the cost. In the circuits to be described the same type of relay as used in the earlier circuits was employed, that is, a $670 \Omega 14 \mathrm{~mA}$ operating type.

## Time Constant Circuit.

The normal method for obtaining the time delay with a transistorised circuit is to wire the relay in series with either the emitter or collector of the
transistor and connect a time constant circuit, which consists simply of a resistor and capacitor in series, between the positive and negative supply lines with the junction of the two components wired to the transistor base.

If the capacitor is wired to the negative line, then the relay will come on as soon as the operating voltage is applied to it but will drop out again after the required time delay: whereas if it is the resistor that is connected to the negative line then the delay will occur before the relay operates.

In practice the input or base to emitter impedance is fairly low, in the order of a few hundred ohms, when sufficient voltage is applied to it to cause a current of several milliamps to flow in the collector circuit. This low impedance acts as a shunt across the time constant circuit, with unfortunate results.

An approximation of the time constant given by a series resistor and capacitor is given by the formula $T=R C$. where $T=$ time in seconds. $R=$ resistance in ohms and $\mathrm{C}=$ the capacitance in Farads. Thus a resistance of $100.000 \Omega$ must be used with a capactor of $100 \mu \mathrm{~F}$ in order to obtain a delay of ten seconds.

If the transistor input impedance shunts the resistor down to a value of only 100 n then a capacitor of $100.000 \mu \mathrm{~F}$ will be needed to give the ten second delay, an impractical value.

Another way of looking at the problem is to say that a certain voltage must be applied to the base in order to cause a sufficient voltage to appear across the relay to make it operate. hut the time constant circuit forms a voltage divider chain to

the base. With the resistor shunted to a value of a $100 \Omega$ or so in the "delay before make" type circuit and using the specified $100 \mu \mathrm{~F}$ capacitor it would probably be necessary to apply over a hundred volts across the chain in order to reach the required base voltage.
The way round this difficulty is to connect the relay in the emitter line, in which case the base voltage is applied between the base and the positive line and not base and emitter, with the result that the input impedance is raised in much the same way as happens with the valve type cathode follower circuit. Roughly speaking the input impedance then becomes the product of the transistor current gain and the resistance in the emitter line, in this case the relay coil resistance.

## Raising the Input Impedance

The input impedance gained in this way may still be low enough to cause severe difficulties, but it can be raised still further using two transistors, as shown in Fig. 3.

Here it can be seen that the emitter of a second transistor. Tr2, is connected to the base of the first and that the collectors of both Trl and Tr2 are connected to the negative line. The operation is as follows:

Assuming the rough calculation of input impedance above to be correct and the gain of the transistors to be 30 . then if a relay with a coil resistance of 1.000 ? is used it is found that the input resistance at the base of $\operatorname{Tr} 1=30.000$ 上. The base resistance of Trl also forms the emitter load of Tr2. so that its input impedance becomes $30 \times$ $30.000=900.000 \Omega$ s or so. In practice impedances of several hondreds of h!2 can be expected. This is sufficiently high to cause negligible shunting to most time constant circuits.

It is found in practice that the time delay required is best found by experiment as it depends not only on the theoretical resistor and capacitor values, but also on the characteristics of the relay. the transistors and on the voltage of the supply.

In the circult shown in Fig. 3a. which is of the "delay before operate" type, it was found that with the relay already stipulated a delay of six seconds occurred using a resistance of 33,000 !2 and a capacitance of 100 aF and a battery voltage of 17 V . The relay is started by switching the supply on to the unit via SI. When the supply is turned off again the unit resets almost at once, a delay of a small fraction of a second occurring while the capacitor discharges itself. The reason for the fairly large battery voltage is that the transistor used would only pass a little over the required 14 m A to operate the relay from a 12 V supply even when the base of $\mathrm{T}: 2$ was connected to the 12 V negative line directly. 17 V is. of coursc, well above the recommended maximum rating for Red Spot transistors and although the writer has used such voltages many times in the past without a singie failure it is possible that break-down may occur in some cases.
It is therefore recommended that in cases where it is decided to use a transistorised delay circuit on a relay which requires an operating current of greater than 14 mA within the voltage ratings of the transistors. Tri be replaced either by an OC72 or OC81. or preferably, with a view to keeping the cost down, by two or more Red Spots in parallel, that is, wired emitter to emitter, collector to collector and base to base.

## On-Delay-Off Circuit

The circuit shown in Fig. 3b is the arrangement for the on-delay-off type of unit. In the prototype circuit a delay of tive seconds was obtained using a capacitor of $100 \mu \mathrm{~F}$ and a resistor of 47,00002 with a battery voltage of 14 V .

It should be noticed that with this circuit once the relay has dropped out the circuit does not automatically reset itself, as the capacitor retains its charge even if the battery supply line is broken. It is thercfore necessary to use a two pole, two way
switch for operating the circuit, with one switch bank in series with the supply and the other wired across the capacitor so that when the power is turned off the capacitor is automatically short circuited and discharged. Resetting then becomes automatic. There may be occasions wher time delays or incrased sensitivity are not required at all, and the only problem presented by the relay under consideration is its wide on/off voltage ratio. As already mentioned, this ratio geverally lies between about $3: 1$ and $4: 1$ which is large enough to present many problems in what can be described as servo-operated equipment where the relay is intended to come on only when a sound, light. heat or other quantity reaches a certain value and to drop out again when the quantity falls below that value again.

## Reducing the On/Off Ratio

A simple solution to this problem is to wire a Zener diode in series with the relay, as shown in Fig. 4. The reduction in the on/off ratio is a function of the ratio of the Zener voltage/relay voltage ratio.

The action of the Zener is to more or less raise the on/oft

Fig. 4: Using a Zener diode to reduce the on/off ratio.
voluges of the rclay by the value of its own voltage. For
 example. in a test circuit it was found that a relay came on at $8 \frac{1}{2} \mathrm{~V}$ and dropped out at $1 \frac{1}{2} \mathrm{~V}$. giving a ratio of about $5 \cdot 7: 1$. a little wider than average. When an 8 V Zener was connected in series with the relay it was found that the on voltage became 15 and the off voltage $8 \frac{1}{2}$, giving a ratio of between $1 \cdot 7: 1$ and $1 \cdot 8: 1$, a great improvement.

In a second test a $1.000 \Omega$ relay with an on voltage of $10 \frac{1}{2}$ and an off voltage of 3 was used. giving an on/off ratio of $3 \cdot 5$. When the 8 V Zener was connected in series the voltages became $16 \frac{1}{2}$ and 10 respectively. giving a ratio of 1.65 .

A third test used a $10.000 \Omega$ relay with on/off voltages of 10 and 3 with a ratio of $3 \cdot 3: 1$. When the 8V Zener was used these became 18 and 10.5 giving a ratio of about $1 \cdot 8: 1$.

The higher the Zener voltage used the smaller becomes the on/off ratio, so that the ratio can be made to be determined only by the maximum available supply voltage and the sensitivity of available relays.

If for example, a relay with on/off valtages of 10 and $2 \frac{1}{2}$ is available. giving a ratio of $4: 1$, and is to be used in a valve circuit where an operating voltage of say 200 may be used, then a number of Zener diodes may be wired in series so give a 7 ener voltage of 190 . It will then be found that when the relay is put in series with them the on/ of voltages will become about 200 and 197 so that the on/all ratio is then reduced to wirtually $1: 1$.

It must be realised. of course, that it is only the on/aff wolfage ration that is altered in this way, the on/offeurent ratios remaining the same.

## A tape auto stop <br> BY K. A. CARR-GLYNN

MANY less expensive tape recorders are not fitted with automatic tape stops to halt the drive mechanism when the tape has run to the end of the spool, but the device to be described enables practically all such tape recorders to be modified for " auto stop".

Most tapes have stop foils at the ends. before the last leader, and this foil is used to actuate the tape stop. The circuit as shown in Fig. 1 is a very simple one, though very effective.

MR1 is a meter rectifier of the bridge variety. though, as it is only the heater voltage that is being rectified and in normal conditions there is no drain on it, there is nothing against the use of germanium diodes here.

When running the tape normally, the negative potential from MRI is applied to both grids of VI, which is consequently cut off. The reason a double triode is used is the need to provide a large current increase to operate the relay RL1. When the valve is cut off the relay is de-energised and supplies are made to the tape amplifier and drive motor. This state is indicated in Fig. 1.

However, when the metal leaders on the tape


Fig 2: A brass shim fitted to a tape pressure guide.
short out the two contacts " $x$ " and " $y$ " the grids are earthed, and a large current flows through the valve and the relay becomes energised. Thus the motor supply is broken. and also. in order not to increase the current drain on the h.t. rectifier, the amplifier supply is broken also.


Fig 3: A suitable arrangement for the 'switching guides'.


Fig. 1: The simple circuit of the device.

It can be seen from this, however, that the beginning of the tape will also stop the tape. and to prevent this action a three-way switch can be inserted, shown as S 1 on the diagram.

When S1 is placed in the "override" position the foil does not actuate the stop device, and when placed in the "stop" position the grids are earthed and the motor and amplifier supplies broken.

In order to operate this device two contacts are needed which will be shorted together by the metal foil. In the Philips recorder which the author possesses, one of the tape guides was a glass tube. This wa, covered with a brass shim and one of the tape pressure guides was used as the other contact as indi* cated in Fig. 2.

Two of these contacts or "switching guides" could be fixed on to the deck atany convenient point (Fig. 3). It is essential to ensure that brass shims are perfectly flat on the glass. The shims should be well polished with grinding paste, then with "Brasso" to ensure a perfectly smooth surface.

## a voice-operinted

This piece of equipment will operate on a baby's cry and will relay the output through the audio stages of a television receiver.

By G. G. Scott

'1HIS alarm device consisis of a voice operated switch and it was designed to work in conjunction with the output stage of a television receiver.

The television receiver audio circuit is broken between the volume control and the grid of the a.f. amplifier. Connection is tahen from the centie sifider on the volume control to a relay which is mounted on the baby alarm unit.

In the de-energised condition of the reldy. the television sound is switched bath 10 its own amplifier. When the relay is energised due to a signal in the alarm, the andio output from the television receiver is switehed off and the atudio output from the baby alarm is switched to the television set.

This syctem was designed as the normal baby alarm with its own output stage and speaker and was found to have cerlain disadvantages.
(1) Difficulty was found in detecting whether the sound was coming from the alarm or the TV set.
(2) If it was a good programme on TV. ne waited a minute or so before seeing to baby, while

if a baby appeared on TV one was always running to see to baby!

## The Circuit

The microphone pre-amplifier is of the simple double tride (ECC83) type, designed more for simplicity rather than yuality.

Two outputs are taken from the second stage, VIB. one via the relay contacts to the televisto, receiver. and the other to a suare wave limiter.

A IMO grid stopper (R10) is used in the limiter stage $V 2$. When the signal on the grid of V 2 is more posisive than the cathode. grid current flows


The finished equipment.
so that the grid only goes slightly positive and is held at a more or less constant potential, thus syuaring the positive peaks. On the negative half cyctes, when the signal goes beyond grid base, tlee valve is cut off thus squaring the negative peaks.

A bleeder network, R8, R9, from h.t. is used so that the cathode potential is less dependent of valve current. It was also found that the screen produced a better square wave than the anode, so the screen is coupled through a 1500 pF capacitor C3 to a double diode, V3.

Fig. 1: The necessary connections from the relevision receiver.

## Operation of Diodes

On the positive half cycle the right hand plate of C 3 follows the square wave output. This makes the anode of diode V 3B positive, and it conducts, so charging up the 0.05 capacitor C4 in its cathode (to approximately 50 V ) as well as the 1,500 pF capacitor C3 (see Fig. 3).

On the negative half cycle the squate wave voltage drops by about 200 V thus the right hand plate of C 3 drops to approximately 150 V negative. As diode V3A cathode is now at 150 V negative with respect to its anode, it conducts thus discharging C3 to its original state. As diode V3B anode is also negative this diode is cut ofli and C4 dnscharges through the 1 M@ resistor R14.

On the next positive half cycle the whole operation staris over again.

The 1 MQ resistor R 15 and the $0.1 \mu \mathrm{~F}$ capactor


Fig. 2: The circuit of the unit.


Fig. 3: This illustrates the operation of the diades (see text)

C5 form a filter so that a d.c. proportional to the frequency is applied to the grid of the relay amplifier V4.

The cathode of the amplifier valve is connected to a bleeder network from h.t. The $25 \mathrm{k} \Omega$ variable VR1 (wire wound) biases the valve to cut-off when no signal is applied. The correct bias is found by talking fairly quietly into the microphone and adjusting VRI until the relay just energises.

As the d.c. voltage on the grid of the relay amplifier is proportional to frequency, then the bias on the cathode can be arranged to suit the frequency required.

The power supplies are derived from a 6 V heater transformer and two RM1 half wave metal rectifiers connected in series.


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RADIO, TELEVISION AND ELECTRICAL REPAIRS
Edited by Roy C. Norris; published by Odhams Books Ltd. 576 pages, 6 in. $\times 9$ in. Price 50 s.

THIS is the fourth edition of a well-known publication. but is by no means a rehash. A team of eight specialist contributors, under the guidance of Roy C. Norris, Technical Editor of "Electrical and Radio Trading ", has completely rewritten the work.

The subjects range from water heaters, domestic appliances and electrical installations, through radio and TV adjustments and repairs to the latest u.f.f. innovations that pave the way to colout and there is even a section specially prepared to explain the various colour IV systems. Also insluded is a comprehensive section on the sevicing of gramophone and tape recorder mechanisms.

Whether the handyman wants information on clectrical wiring. bench work, power supplics or refrigeration, or the practising engineer requires to biush up his knowledge of current techniques of radio and television repairing, an authoritative reference will be found in this comprehensive volume.

There are thirty chapters, each a dozen pages long, an initial table of abbreviations and symbols and an index that runs into five pages of small print. In the 576 page work there are no less than 470 illustrations.

It is perhaps on this last point that the reviewer might find cause to carp. There is a curiously oldfashioned flavour about the drawings and their presentation. No photographs have been used. The very annoying habit of omitting component values from circuit diagrams has been retained, except in the fault-tracing section.

But this is a minor defect in an ambitious publication that should not be absent from the bookshelf of anyone, amateur or professional, who is concerned with radio, television or electrical repairs.-B.R.G.

## SHORT WAVE AMATEUR RADIO

By 1. Schapp, PAOHH. 160 pages, $5 \frac{2}{4} \times 8 \frac{1}{4} \mathrm{in} .7$ fold-out diagrams, Price 21 s.
Published by Philips Technical Library. U.K. Distributors: Cleaver-Hume Press Ltd.

'ITHIS book is not a handbook which could be described as essential for experienced amateurs, nor is it a basic book for the novice, but tends to steer a middle course. Unfortunately, in a small volume such as this it does. of necessity, only offer a very brief discustion on the various technical aspeets of amateur radio.

It omits any great discussion of single sideband. a mode which is becoming increasingly popular, and also makes no attempt at v.h.f.. this being described as outside the scope of the book. It is felt that this hook has a limited appeal on the British market although had values been included
with all the circuits shown it might well have proved more useful. The preface warns that this is not an extensive practical manual and the brevity with which it deals with some topics may be judged by the fact that transmission lines are dealt with in two pages.

At the back are pull-out circuit diagrams of two transmitters and two receivers. The wire for the coils is not specified in s.w.g.. but in actual diameter in millimeters. Although there are no layout diagrams and few constructional details there are photographs of the equipment, which would prove handy to the would-be constructor.

There are a number of useful appendices contained in the last twenty-eight pages. Included amongst these are bandspread calculations, coil calculations. amateur abbreviations, and an ahridged version of the list of countries prefixes. D. (i.

## DIODE CIRCUITS HANDBOOK

By Rufus P. Turner. 128 pages, $8 \frac{1}{2} \times 5 \frac{1}{2}$ in. Price $\| 8 s$.

## INTRODUCTION TO ELECTRONIC SERVICING

By J. A. Stanley. 160 pages. $8 \frac{1}{2} \times 5 \frac{1}{2} \mathrm{in}$. Price 24s.
Eoth books originally pubiished by Howard W. Sams \& Co. Inc., Indiana, U.S.A. Published and distributed in the U.K. by W. Foulsham \& Co. Ltd.

1YHE diode is perhaps one of the simplest components to be found in the world of wireless and electronics. So simple. in fact, that many people tend to regard it as merely a glorified cat's whisker. It"s the "small thing with two wires", and asked to name some uses of a diode. the average reader might reply; "rectification, detec-' tion and . . .?"

For these who. like the writer, have sadly underestimated the humble diode a book has been written and in it Mr. Turner describes nat two, but nearly a hundred uses for a diode. Not only describes them but furnishes with each a practical circuit diagram complete with values. Still not convinced? Then, let's flick through a few pages entirely at random. and see just how versatile and useful these "things with two wires" are:-

Page 35 shows two circuits, one of a c.w. transmitter and the other a 'phone transmitter. The valve line up-one diode. A diode capacitor memory cell is depicted; again only one diode is necessary. One could jump about the book and name circuits to suit all takes. The photo diode, tape or punched card reader, a "grid" dip oscillator (yes only one diode for this too) an a.f. milliameter, a diode matrix to name just a few.

In case some readers are still thinking in terms of "cat's whiskers". Mr. Turner includes many circuits which use the newer Zener diades, tunnel diodes, and a newcomer called the Varactor. at device whose capacitance is dependent upon the boltage applied to it. A snag which maght arise is the avalabilaty in this country of the particula!
continued on the following pcge

# A SIMPLE R.F. AMPLIFIER 

THIS unit is primarily intended for use with those domestic receivers which cannot be provided with an efficient outdoor aerial. It consists of an untuned radio frequency amplifier which can appreciably improve the medium and long wave performance of the usual four stage type of superhet receiver. The amplifier will also prove to be useful for most crystal sets and t.r.f. receivers.

The circuit is exceptionally simple and construction should therefore present no difficulty to less experienced constructors. Of course the usual care must be taken when soldering the transistor and when connecting the battery since errors will almost certaimly destroy the transistor.

## Economical Power Consumption

The circuit is very modest with respect to power supplies and works efficiently when taking a current of about 1 mA from a $4 \frac{1}{2} \mathrm{~V}$ torch battery. With long and therefore more efficient aerials the amplifier may tend to overload the receiver when it is tuned to powerful stations. It is thus necessary to be able to adjust the gain of the amplifier and this is achieved by reducing the current flowing through the transistor by means of VR1.

## Connecting to the Receiver

With most receivers there will no difficulty in connecting up the amplifier. However very great care must be taken when dealing with receivers of the a.c.-d.c. (all mains) type. In such cases ensure that the neutral wire of the mains is connected to the chassis and also that the earth and aerial sockets of the receiver are isolated from the chassis via suitable capacitors. It is also advisable to dis-

BY J. ROBERTS


Fig. 1: The simple transistor circuit.
connect such receivers from the mains supply when connecting the amplifier.

## Performance

It is difficult to give an exact idea of the improvement likely to be noticed, as so much will depend on the particular receiver and aerial in use with the amplifier.

The amplifier has been tested with receivers other than the domestic type and this example of a test on a communications receiver may be of interest:

National 9 Valve superhet. 8 ft . indoor aerial. time 11 p.m., medium wave band: 26 stations readable without amplifier: 51 stations readable when the amplifier was added. The results of this test were checked carefully and there is no error.

## BOOKS REVIEWED

## -continued from previous page

diodes indicated since this book is of American origin. Any reputable stockist would perhaps advise as to British equivalents, and of course, it is possible to send to the U.S.A. for the diodes with the happy knowledge that postage would be reasonably cheap.
Other than bearing this one point in mind the book would appear to be a useful addition to any Library. Since the majority of the circuits contain only one or perhaps two diodes the book should appeal to experimenters with a limited budget. Indeed, there is a separate chapter entitled "Experimenter's Circuits", which would be ideal for readers whose interests lie in this field.

WRITING a book strictly for a novice is no light task, especially when the subject is on a highly technical topic such as a branch of electronics. It is one thing for an author to be knowledgeable and highly skilled, but to write clearly and concisely so that a raw beginner will fully understand is a very different kettle of fish.

In " Introduction to Electronic Servicing for the Begimer", Mr. Stanley has proved beyond any shadow of doubt that he is able to write with great clarity and simplicity and in such a way thit dav novice will find it a very difficult task not to understand.

This book is thoroughly recommended and is excellent in every way. It takes the novice gently by the hand and guides a good reliable solid path through both theory and construction. It is profusely illustrated. and gives both circuit diagrams and actual drawings. When a new section or extra components are added, the existing wiring is shown "ghosted" and the new additions are shown clearly in black. It teaches how to read a circuit diagram, and by the excellent methods mentioned gives full details for the construction of a power unit, signal tracer, multimeter, and superhet tuner, together with full details of how to use, how to test, and how to repair. Although the price is 24 s .. there can be little doubt that it is a bargain.

The book is of American origin. but no risks are taken due to differences in electronic practice, and a two and a half page introduction is especially written for the benefit of British readers.

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# TEST GEAR accessories 

## Part 6

BY C. MACKAY

cision shunts and droppers, can give a result of as high an accuracy as one's good (say 20,000 s per volt) multimeter-provided that care is taken to use this second meter in such a way that it does not noticeably affect the circuit conditions.

The circuit of a suitable multimeter built by the author is given in Fig. 22a. The meter used was of the 1 mA f.s.d. type with an internal resistance of $100 \Omega$.

If a meter with any other value of internal resistance is employed R7 becomes $\frac{\mathrm{Ri}}{99}$ and R 8 becomes $-\frac{\mathrm{Ri}}{9}$.
Seven ranges are available: 0-1mA d.c. $0-10 \mathrm{~mA}$ (.c., $0-100 \mathrm{~mA}$ d.c.. $0-1 \mathrm{~V}$ d.c. $0-10 \mathrm{~V}$ d.c., $0-100 \mathrm{~V}$ d.c., $0-1,000 \mathrm{~V}$ d.c.

## CONTINUITY/INSULATION TEST

No facilities are included for the measurement of a.c. voltages and currents as the "good" multimeter i* generally adequate.

As shown above, a range for testing insulation and continuity is provided, though no separate calibration scale was given for this range. However, if required it could be calibrated as in Fig. 22b.

Good quality $1 \%$ tolerance components must be used for R1-R4, R7 and R8.

In order that the overall accuracy of the completed instrument should be maintained as high as possible the author advises that the stunts (R7 and R8) should be bought and not made up. These can be purchased at a reasonable price from the Planet Instrument Company, which advertises

| R1 R2 | IM $21 \%$, IW $100 \mathrm{k} \Omega 1 \%$ |
| :---: | :---: |
| R3 | $10 \mathrm{k} \Omega 1 \%$ FOR FIG. 22a |
| R4 | lks 1\% |
| R5 | 1 k 12 wirewound |
| R6 | 3,900s2 10\% |
| R7 | $1 \cdot\|\mid 182 \%$ wirewound |
| R8 | 11-1\|s2 $1 \%$ wirewound |
| MI | Moving coil meter. ImA f.s.d. Internal resistance $100 \Omega$ |
| SI | 1 pole, 8 way, 2 bank wafer switch |
| B1 | 4.5 V flat battery |

## Miscellaneous:

Two terminals, one red, one black.
Fig. $22 a$ (left) : A multimeter circuit as used by the outhor; $b$ (below): calibrations for an ohms scale.

Miliamperes

regularly in this magazine. R 6 is for adiusting the meter to read full scale on the resistance range with the input terminals shorted out.

## mounting the meter in its case

The case used by the author was a rejuvenated wooden box used for protecting a large mains transformer in transit. Virtually any case, wooden, plastic or metal, can be used. If it is metal the terminals, which can be spade terminals or wander plug sockets, should be well insulated from the box. The matter of the cabinet will therefore be left to the ingenuity of the experimenter.

A few suggestions worth some consideration are using biscuit or sweet tins, lanch boxes (metal or plastic), Oxo cube boxes or baking tins with a separate aluminium baseplate.

This unit can be easily and completely constructed for under $£ 2$.

## USING THE "SECOND" METER

If this meter is placed across a circuit wherein the current flowing is of the same order as that which will be taken by the meter itself, the extra load imposed upon the circuit by the meter will completely change the circuit conditions and all readings taken by the meter will be inaccurate if


Fig. 23 (a): An oudio amplifier; (b) measuring with a $20 \mathrm{k} \Omega / \mathrm{V}$ meter; (c) with a $1 \mathrm{k} \Omega / \mathrm{V}$ meter.
not completely false.
A meter which draws only say, $50 \mu \mathrm{~A}$, therefore. will not have nearly so drastic an effect as will one which draws 1 mA .
This can be seen more clearly in Fig. 23a. Here, a valve, V1 is drawing 0.5 mA from the h.t. + supply. In Fig. 23b a 20,000 s per volt meter is connected across point $A$ and chassis in order to measure the anode voltage. This meter will draw only 0.05 mA from the circuit and the error of the reading will be only because the current has increased from 0.5 mA to 0.55 mA .

In Fig. 23c, a $1,000 \Omega$ per volt meter is connected between $A$ and chassis. As this meter takes its maximum current of 1 mA , the anode voltage will


Fig. 24: "Cause and effect measurements" where a pair of meters is indispensoble.
be greatly reduced (owing to the increased current through R1) and the error will be so great as to make the reading quite useless.

The reader by now, will probably be wondering why a meter of this sort has ever been used at all. The reason is that these meters can be used for voltage measurements where the current taken by the meter does not affect the performance. They may also be used for current measurement so long as the current to be measured is high enough to give a readable reading (tusually $5 \%$ of f.c.d. is enough).

## TWO METERS INDISPENSABLE

An example of a circuit test where two meters are indispensable is given in Fig. 24. Here, it is desired to measure the amplification of a triode valve with a given anode load. The "good" meter, which draws a low current, is used to measure the anode voltage, while the low sensitivity meter is used to measure the invut voltage, which is adjusted by means of R1, a potentiometer attenuator.

The low sensitivity meter is also highly satisfactory for the measurement of current. For example. this meter could have been used to measure the anode current accurately in Fig. 23. As the anode load resistance would only be increased by perhaps one part in two or three thousand. the results obtained with the meter in that position would be most accurate. At the same time the "good" meter could be used for voltage measurement on the anode as in Fig. 23 b.

This relatively simple meter will be found to be invaluable to the experimenter if it is used with only a little forethought. The instrument is of a well-tried and thoroughly reliable design and will give many years of reliable service.


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Availables accessories include a 2500 V d.c. multiplier and 5, 10 and 25A shunts for d.c. current measurement.



The instrument is supplied in an attractive black carrying case, which also houses a pair of leads with interchangeable prods and clips. and an instruction booklet. It is packed in an attractive display carton. Robust real leather cases are available, if required, in two sizes. one to take the instrument with leads, clips and prods, and the other to house these and also a high voltage multiplier and a d.c. shunt.

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ACTON, BRENTFORD AND CHISWICK RADIO CLUB Hon. Sec.: W. G. Dyer, G3GEH, 188 Gunnersbury Avenue, Acton, London, W.3.

The Club next meets on 9th June when members will review the experiences and results of their participation in National Field Day.

## AMATEUR RADIO CLUB OF NOTTINGHAM

Hon. Sec.: R. W. Attenborough, G3RDJ, Beech House, Chapel Lane, Epperstone, Nottingham.
The meerings of this Society are held every Tuesday evening at the Sherwood Community Centre, Woodthorpe House, Mansfield Road, Sherwood.

## BRADFORD RADIO SOCIETY

Hon. Sec: E. G. Barker, G3OTO, 63 Woodcot Avenue, Baildon, Shipley, Yorkshire.
Members enjoy a full programme of events for May, beginning with a discussion on N.F.D. arrangements on the 12 th. In the same month a party from the elub were the guests of the Spen Valley Radio Society, and on the 26th, J. Spivey (G2HHV) gave a lecture called "V.H.F. Operating.'
CHESTER AND DISTRICT AMATEUR RADIO SOCIETY Hon. Sec.: P. J. Holland, "Field House," 19 Kingsley Road, Gt. Boughton, Chester, Cheshire.
Members attended the meeting for May 5th to hear a lecture. A week later, members were invited to bring any valves with them for testing on the valve-tester which G3EWZ had brought along for a demonstration. National Field Day arrangements were under discussion on May 26th.
IPSWICH RADIO CLUB
Hon. Sec.: R. J. Wells, 43 Clench Road, Holbrook, Ipswich, Suffolk.

This society's activities cover all aspects of radio and therefore any potential members are welcomed to the meetings which are held at lpswich Civic College on the last Wednesday in each month, commencing at 7.30 p.m.
LICHFIELD AMATEUR RADIO SOCIETY
Hon. Sec.: V, Hickman, G3LXR, 143 Main Street, Stonnall, Walsall, Staffordshire.

The membership of this Society continues to grow with more and more licensed amatours among its numbers. One honorary member, T. Painter (G3NEU), has recently presented the club with an inscribed silver cup which will become the trophy for forchcoming d.f. events.
MELTON MOWBRAY AMATEUR RADIO SOCIETY
Hon. Sec.: D. W. Lilley, G3FDF, 23 Melton Road, Asfordby Hill. Melton Mowbray, Leicestershire.

The only club event for May was on the 215 t when J. L. Warrington (G2FNW) gave a lecture entitled " 70 cms ."
NORTHERN HEIGHTS AMATEUR RADIO SOCIETY Hon. Sec.: A. Robinson, GJMDW, Candy Cabin, Ogden, Halifax.

Sociery events for May began with a junk sale on the 13th. A fortnght later Mr. L. L. Cobb (G3Ul) gave a talk on "Transistors." PRLSTON AMATEUR RADIO SOCIETY
Hon. Sec.: W. K. Beazley, G3RTX, 9 Thorngate, Penworth. am, Preston, Lancashire.

On May 12th, a group of members enjoyed an interesting visit to the Ribble Generating Station. Later in the month, on the 26th, the final arrangements for N.F.D. were discussed.
READING AMATEUR RADIO CLUB
Hon. Sec.: R. G. Nash, G3EJA, "Peacehaven," 9 Holybrook Road, Reading, Berkshire.

Members G2FZI and G3EJA have both donated cups and T. Bingham a shield for Club trophies to be awarded to the winners of contests held during this year. The meeting for May 30th was devoted to a sale of surplus equipment.
SALOP AMATEUR RADIO SOCIETY
Hon. Sec.: Dr. K. E. Jones, G3RRN, Greystones, Shrewsbury Road, Church Stretton, Shropshire.

This Society reports an increasing membership with very good attendances at the meetings.
SCARBOROUGH AMATEUR RADIO SOCIETY
Hon. Sec.: P. B. Briscombe, G8KU, "Roseacre," Irton, Scarborough, Yorkshire.

A sale of members' surplus gear was held on May 7th. A talk on the SXIll receiver on May. I4th was followed on the 21 st by a

display and demonstration of home constructed equipment. On May 28th, members attended a talk given on "Oscillwscopes." SOUTH SHIELDS AND DISTRICT AMATEUS RADIO CLUB
Hon. Sec.: D. I. Forster, G3KZZ, 41 Marlborough Street, South Shields, Co. Durham.
Members from this Society were among the visitors to the mobile rally held at Harewood House, Yorkshire, on 24th May. The Club is organising its own mobile rally on 5th July to be held at Bents Park Recreation Ground, Coast Road, Sosth Shields. at Bents Park Recreation Ground, Coast Road, Sonth
SPEN VALLEY AMATEUR RADIO SOCIETY
Hon. Sec.: N. Pride, 100 Raikes Lane, Birstall, Leeds.
"Missilemen 1964" was the title of the lecture given to members of this Society at their meeting on May 14th by Mr. M, A. Browne. On May 21st. members paid a visit to Research Electronics at Cleckheaton and on the 27 th , David Pratt (G3KEP) tave a talk entitled "RTTY." Another visit was organised for the first event in June, when a group of members went to the Met. Office at Church Fenton.
WESSEX AMATEUR RADIO GROUP
Hon. Sec.: P. Cutler, 43 Langside Avenue, Wallisdown, Poole, Dorset.
This Sociery has recently decided to increase its meetings to two a month and also increase its amateur radio activities. To put into effect this latter decision, the Group elected to purchata a 20 m s.s.b. transceiver and put 2 m station on the air.
WIRRAL AMATEUR RADIO SOCIETY
Hon. Sec.: A. Seed, 31 Withert Avenue, Bebington, Wirral, Cheshire.
A mobile week-end in Wales was organised by member G2MMV and was held successfully during May.
WORCESTER AND DISTRICT AMATEUR RADIO CLUB Hon. Sec.: G. W. Tibbetts, 35 Perdiswell Park, Droitwicb Road, Worcester.
Activities for May consisted of a junk sale and practice run for National Field Day. The sale was held on the 16ch and the practice run on the 24 th .
SEVENTH LONGLEAT MOBILE RALLY
Rally Organiser: E. C. Halliday, G3JMY, 4 Parkide Avenue, Winterbourne, Bristol.
This year's mobile rally at Longleat, near Warminger, Wiltshire, is to be held on June 2lst and talk-in-stations wifh be G3JMY/A on top band and G3SJl'A on 2 m . The entrance to Longleat House park and estate lies on the Frome-Warminster Road (A362). There will be the usual attractions; displays. comperitions, demonstrations, etc, and once again this event promises to te a memorable occasion.

## .. AND FOR YOUR FURTHER EN JOYMENT

 why not get the June issue of PracticalTelevision, on sale now?
Featured in this issue is a C.R.T. Tester and Rejuvenator.

Also details for converting receivers for reception of Continental programmes, and many more interesting articles and regular features.


New Portable from Roberts Radio
A NEW transistor portable, Model R. 300 has recently been announced by Roherts Radio Company Limited, Molesey Avenue, West Molesey, Surrey. This set covers medium and long waves and incorporates six transistors and three diodes.

A coaxial socket for a car aerial is included in the design and a carrying cover made of strong grey waterproof material is available as an extra.

The size of the R. 300 is $4 \frac{1}{\frac{1}{2}} \mathrm{in}$. $x 5_{\frac{7}{x}}^{2} \mathrm{in}$. $x 4 \frac{1}{\frac{1}{2}} \mathrm{in}$., and its price is 15 gns. Power is provided by an Ever-Ready PPY battery, and current comsumption with no signal is 12 mA .


The R. 300 portable receiver, new from Roberts Radio

## The "Audio Heart"

MARKETED by Technical Suppliers Lid., 63 Goldhaw Road. London, W'.12, is a miniature high quality transistor amplifier designed to be used at the centre of any good audio assembly. The "Audio Heart" gives IW undistorted output for an input voltage of 5 to 10 mV . Input impedance is 1.5 to $2 \mathrm{k} \Omega$. and output impedance 3 to $5 \Omega$. The operating voltage is 6 V , and the signal-tonoise ratio is said to be below measurable level.

Among its numerous applications, the "Audio Heart" can be used with a radio tuner for use in a car, as a burglar or baby alarm, etc., and two of
the units can be used for stereo applizations.
The size of the "Aladio Heart" is 3 in . x 2.2in. $x \quad 1 \cdot 2 \mathrm{in}$. high. and it is atailable ready built and tested at a price of 59s. 6ul.. or in kit form with full building and operating instruetions for 52s. 6d.


This is the "Audio Heart" marketed by Technical Suppliers Ltd.

## Two New Receivers

( ${ }^{\text {RUNDIG }}$ (Great Britiain) Lid., announce the introduction of two new portables to their range of receivers. They are the "Music Boy 204 ", and the "City Boy 204". The Music Boy has a bandspread 49 m band and is claimed to give crystal clear reception over a wide range of European short wave stations. Other bands covered are medium, long and v.h.f./f.m.
The circuit inchides nine transistore, three diodes and one rectifier. Power output is IW and power is provided by two 4.5 V flat torch batteries or a y power pack. The speaker is an elliptical 4in. $x$ Gin. model fitted with a ceramic magnet.

The Maric Buy weighs approximately 4lbs. without batteries and measures 1 lin. $x$ 6 $\frac{1}{2}$ in. $x$ $23^{\circ} \mathrm{in}$. Price is 39 gns .

The "City Boy" embodies nine transistors, three diodes and one rectifier. and corers medium. hang and v.h.f. hands. Output is IW and this is fed into a oin. x 4 in . elliptical speaker. Weight is approximately 3! Jhe. without batteries and the dimensions are $10 \frac{1}{4}$ in. $\times 6 \frac{3}{3}$ in. $x 2 \frac{5}{8}$ in. Price is fixed at 34 gns.

Grundig (Great Rritain) Ltd., Newlands Park, Sydenham, Loulon, S.E.26.


The Grundig "Music Boy 204",

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I.50. 50 watts Hi-Fi Ultra Linear. size $14 \times 10 \times 81 n$. high. £23.2.0. Cover with carrying handles $32 / 6$.
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Adding extra ranges to simple meter movements is one means of making

## inexpensive multimeters <br> \section*{CONTINUED FROM PAGE 149 OF THE JUNE ISSUE}

LAST month some general hints were given on choosing a milliammeter and converting it to a multi-range instrument. In this month's article the conversion of a piece of surplas equipment containing a suitable meter is described.

The instrument described here was made from a piece of "surplus" equipment which rejoices in the name "Gear, Electric Firing Mark Two: Instrument, Testing, Mark 2 or 2 A ". It consists of a centre-zero milliammeter in a wooden case with an ebonite front panel.

The case contains one or two other components and in fact it turns out to be a Wheatstone bridge reading $0-25 \Omega$, obviously intended for checking the continuity of detonating circuits of some kind. The circuit is shown in slightly simplified form in Fig. 4.

If anybody requires a single-range bridge for measuring low resis-


Fig. 4: The basic circuit (less the press-switch contacts) of the surplus unit referred to in the text. tances the instrument may be used unmodified. (The average multimeter tends to be inaccurate below about $10 \Omega$. One often wants to measure the resistance of transformer windings, such as the secondaries of output transformers. which may only be a few tenths of an ohm. For such measurements the unmodified unit would be a useful addition to the laboratory.)

However, the writer wanted a cheap multirange meter. This surplus unit cost about 15 s . and the finished article cost about 25 s . The ranges in the writer's version are:
Current $\ldots \ldots . .0-3,0-30,0-300 \mathrm{~mA}$
Voltage $\ldots \ldots .00-3 \mathrm{~V}, 0-30 \mathrm{~V}, 0-300 \mathrm{~V}$
Resistance $\ldots \ldots .0-25,0-250,0-2,500 \Omega$

## Basic Information

Several of the units in the surplus dealer's shop had bent meter pointers. Evidently the sappers who had used them had disregarded the warning label telling one not to switch the meter into
circuit with the firing dynamo going.
The dealer had affixed a label giving the sensitivity of the meter as "600-0-606 4 A. Examination of the meter itself revealed a rated f.s.d. of $660 \mu \mathrm{~A}$. Subsequent measurement revealed that the sensitivity was, in my particular specimen, $647 \mu \mathrm{~A}$. (The sensitivity was slightly different in the two directions of deflection: $547 \mu \mathrm{~A}$ is the average.)

The meter scale reads $30-0-30$ and has 30 small divisions each way. This means that, unless


Fig. 5: Circuit orrongement for finding the coil resistonce of a meter.
one rewrites the scale, the smallest f.s.d. current which fits the scale is 3 mA . The damping of the basic movement was rather poor but good enough to permit an f.s.d. of 1.5 mA . which is posibic without rewriting the scale, at the expense of some simple mental arithmetic.

## Meter Coil Resistance

A test with a $10 \Omega \pm \mathbb{1}$ the original Wheatstone bridge was quite accurate. This suggested the possibility of using the instrument to measure its own resistance, which was shown by a rough check (as described last month) to be within the range of measurement of the bridge.

The method of measuring the coil resistance is of quite general application. What one does is to adjust a variable resistor to the same resistance as the meter and then measure the variable resistance. The technique is illustrated in Fig. 5. Here R1 is any low-resistance potentiometer (say. less than $1,000 \Omega$ ) and R2 is selected to permit f.s.d. with the available voltage.

With the switch open. adjust R1 for f.s.d. Leave R1 alone, close the switch, and set R3 so that the meter reading is, exactly halved: R3 is now equal to the meter resistance and can be taken away and measured.

In the present case one takes the meter out of its original circuit and then, once R3 is set, one puts the meter back again so that the bridge can be used to measure the resistance. In the case of the writer's instrument the meter coil resistance turned out to be $23.3 \Omega$.

## General Design

All the necessary information for designing voltage multipliers and current shunts had now been obtained. The next step was to decide how to use them to obtain a multirange meter. There are two basic ways of doing this.

One is to connect all the multipliers and shunts permanently, as in Fig. 6, and select the ones needed by plugging the meter leads into the appropriate sockets. The other is to use only one pair of terminals and to switch the appropriate shunts and multipliers into circuit. (Combinations of the two systems are possible.)

## Current Ranges

It was decided to use switches and only one pair of terminals. The first problem was to choose between the alternative ways of obtaining different current ranges shown in Fig. 7.

The simplest method is shown in Fig. 7a.



Fig. 6: A multi-range instrument using multiple terminals but no switches.

Fig. 7: Two arrangements for incorporating the current shunts.

It is simplest in that each current range is completely defined by one resistor. Thus R1 might set the 3 mA range, R 2 the 30 mA range, and so on.

The use of this system is often deprecated for the following reasons. First, and most important, the contact resistance of the switch affects the accuracy. The effect may be very serious on highcurrent ranges. For example, the shunt resistance for, say, a 3 A range may be only a few hundredths of an ohm. Even if the switch contact resistance is only a hundredth of an ohm it has an
appreciable effect. Secondly, safety considerations impose limitations on the type of switch. It must have make-before-break contacts. Otherwise, when changing from one range to another, the full current goes through the meter coil.

The arrangement of Fig. 7b avoids both these difficulties. Variations in contact resistance have no effect on the way the current divides between shunts and meter, and during changeover the current is simply switched off. Disadvantages are a higher voltage drop on the higher current ranges and the fact that since all resistors are in circuit all the time, an error in one of them affects all the ranges (though not necessarily very much).

In the writer's opinion there is not much to choose between the two arrangements when only moderate currents are to be measured. Make-before-break switches are easily obtained and contact resistance is quite low with modern precious-metal-plated contacts.

The second system was adopted as a matter of convenience. The meter case of the "surplus" unit is rather shallow and only a single-wafer one-pole ten-way range switch could be accommodated: the second arrangement proved to be easier to use with such a switch.

## Calculating the Shunts

For a Fig. 7a type of circuit calculating the shunts is easy. The voltage drop across any shunt is the same as the voltage drop across the meter. Knowing the resistance of the meter and the f.s.d. current, the voltage drop at f.s.d. is calculable. Thus a $30 \Omega$ meter with f.s.d. 1 mA has a voltage drop of $1 \mathrm{~mA} \times 30 \Omega=30 \mathrm{mV}$. To make the meter read 10 mA f.s.d. we have to arrange that 9 mA flows through the shunt when this voltage appears across the meter circuit. Thus the shunt resistance must be $30 \mathrm{mV} / 9 \mathrm{~mA}=3 \cdot 33 \Omega$. For a 100 mA range 99 mA must go through the shunt, so its resistance R100 (say) $=30 \mathrm{mV} 99 \mathrm{~mA}=0.33 \Omega$. In general if the f.s.d. has to be multiplied N times by the shunt. then,

$$
\text { Shunt resistance }=\frac{\text { Meter resistance }}{(\mathrm{N}-1)}
$$

The Fig. 7b type of circuit is more difficult hecause the voltage dron across the circuit varies from range to range. With the switch in position 1 the drop across the meter is the same as across the three shunts in series. This enables us to calculate the total shunt resistance from the formula above:
$\mathrm{R} 1+\mathrm{R} 2+\mathrm{R} 3=$ Meter resistance $(\mathrm{Rm})$
( $\mathrm{N}-1$ )
But what happens in position 2? If we make the voltage drop across the circuit the same as the f.s.d. drop the meter will read less than f.s.d., since R1 is now in series with Rm. And in position 3 R1 and R2 are both in series with Rm. Thus the circuit drop at f.s.d. increases as higher current ranges are switched in.

Fortunately the various values can be arrived at by a process of elimination as follows:

Step 1. Total shunt resistance

$$
\mathrm{R} 1+\mathrm{R} 2+\mathrm{R} 3=\frac{\mathrm{Rm}}{\mathrm{~N} 1-1}
$$

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| L4 | $2 / 3$ | 6156 | $9 / 6$ | بHW6 | 916 | 311．1．12 11／－ | A741 6／6 | E15 | 4／6 | HM71 | 13／6 | 87 | 281. | TUD4 |  | $\mathrm{F}^{2}$ | 1／6 | cexed 11／8 |
| 1 Lab | 16／10 | 6r ${ }^{\text {d }}$ | $9 / 6$ | 912 | 3）－ | 311.15 | 1：3ik $4 / 2$ | H23 | 12／6 | k3su | 618 | ¢100 | $28 / 5$ | This | 15\％ | Vplo | 16 | GEX 66 15／0 |
| 11.85 | 4／－ |  | $5 / 3$ | 9117 | $7 / \mathrm{E}$ | 201．15 $8 / 8$ | $131.6310 / 6$ | Hiss | 8／－ | EMM |  | ${ }^{2} 338$ | 15\％ | ＇11214． | 10／4 | $1^{1 / 4}$ | 14／6 | SATH00 $/ 9$ |
| $11 . \mathrm{NS}$ | $4 / 6$ | －riad | $3 / 9$ | 161） | $8 / 3$ | $301412 / 3$ | （1）12／6 |  | 4／6 | EMc4 | 1／8 | P61 | $2 / 6$ | 7 T 309 | 14／6 | ${ }^{1} \mathrm{P} 4 \mathrm{~A}$ | 14／6 | MAT1618／6 |
| iNout | $8 / 8$ | $6{ }^{\text {roba }}$ | 7／6 | 10\％ | $12 / 8$ | 30P12 $2 / 6$ | $\mathrm{ClC}^{12 / 8}$ | Ecisa | $27 / 1$ | EMx | $8 / 8$ | Pabct | $18 / 8$ | TH233 | L／8 | $\mathrm{VF}^{4} 4 \mathrm{~B}$ | 20／5 | a1aticu ${ }^{\text {a／g }}$ |
| 111 | $5 / 9$ | br | 31. | 10101 | \％ | 3irlti | 1C1355 12／6 | 4，\％ow | \％ 18 | EMMA | 76 | Pr－xt | 1015 | T102 | 5／－ | VP136 | 7\％ | MAT1\＃1 8／6 |
| 1810 | $4 / 6$ | －F11 | 17／8 | 141） | 11／6 |  | CKбw $0 / 6$ | H2． y ： | e／ | $\mathrm{EN3}$ | 101－ | 1rx | 14／\％ | TP45 | 5／－ | YP23 | 2／6 | OAS 8／－ |
| 1P11 | $8 / 8$ | 61412 | $3 / 9$ | 10 | 10\％ | 30以1 $8 / 6$ | －1．4 19／6 | 41431 | $7 / 3$ | Ex91 | 5／6 | P95 | $11 / 8$ | T12820 | $017 / 6$ | Yr41 | $5 /-$ | Uaila e／－ |
| ${ }^{\text {J H }}$ | 4／－ | OF13 | $4 / 9$ | fors | 9／8 | 301 1 13 976 | 1－1．43 11／6 | Arsis | 4／－ | EYot |  | P＇y7 | $7 / 5$ | Tystif | 11／8 | VP1s3 | $9 / 9$ | טA7t 3／m |
| 18． | $5 /-$ | 6 Fl | 23／3 | 101．14 | 101 | 30 PLI4 12／6 | $2 / 6$ | 1．1434 | $21 / 2$ | 61）1 | 73 | Prest | $5 / 6$ | 1 Alscr | $808 / \%$ | YR76 | 21／－ | OAza 31－ |
| 185 | $9 / 6$ | 8 Fl 15 | $8 /$ | 101.18 | $8 / 3$ | 35 As $20 / 8$ | CV63 10／6 | RCY\％ | \＄10 | HY\％s | 9／8 | Pcess | 6／9 | It $\mathrm{F}_{\text {¢ }}$ | $1=$ | Yelus | 5／6 | vais 8／－ |
| 12－2 | $29 /$ | 8F10 | $6 / 9$ | 101．1）1 | $9 / 6$ | 351．6CT $6 / 8$ | －Vso 14／6 | Eictu | 71 | Y 8 | 9／6 | Pceso | 10／6 | C134 | 10／4 | YR150 | 419 | －481 3／－ |
| 174 | $2 / 3$ | 8 F 17 | $12 / 6$ | $10{ }^{1} 13$ | $8 / 8$ | 35 W4 4／9 | Cve71 12／6 | Ec¢ ${ }^{\text {d }}$ | $3 / 6$ | Eing |  | Pciss | 78 | UBCA1 | 18／3 | vTera | 7／－ | Oass 3／－ |
| 104 | $5 / 8$ | ${ }^{6} \mathrm{FL} \mathrm{\prime}$ | $18 / 5$ | $161^{1 / 4}$ | 11／6 | $35 Z 3{ }^{16 / 4}$ | ＇Y1 16／4 | Farxt | 4／t | ci） 88 | $8 / 8$ | Pcelng | $10 / 8$ | UBCss | ＊／8 | VTS01 | 9／－ | 0．4\％ $4 /-$ |
| 145 | 5／3 | 6F14 | $4 / 9$ | 1113 | $17 / 8$ | 3z\％at $4 / 6$ | （16）6／8 | Ruxa | 4／6 | E）${ }^{\text {c }}$ |  | Prys | 518 | cersu | $3 / 9$ | vell | 51－ | uago 3／－ |
| 287 | 12／6 | 6゙せ3 | $8 / 3$ | 111\％． | 17／6 | 35\％6019 5／9 | （＇Y31 5／8 | Ectay | 5／9 | E／230 | 216 | Prym | 618 | DBFsy | 8／9 | Yと120 | 101－ | vay 3j－ |
|  | $2 / 8$ | 615 ${ }^{\text {ch }}$ | $9 / 6$ | H上1 | 151－ | $3{ }^{3} 1816$ | 1） $1 / 3$ | kres | ${ }^{6 / 8}$ | ER20 |  | Pr PM | $8 / 6$ | Lbler | 9／9 | V1－20a |  | OAys 8／6 |
| －2130 | \％1－ | －524 | 81 | 11 E， | 17\％ | hunda 6／6 | 1151516 | LCcas | $8 / 6$ | E24， | 6／－ | Prexa | 78 | 188 | 6／3 | II 133 | \％ | OASLG 8／8 |
| 21） 21 | 5／6 | bF30 | $3 / 0$ | $1 \because A 5$ | $2 / 8$ | ＋1ata $15 \%$ | 144 $10 / 6$ | ECal | 3／－ | ERas |  | Pr risos | 5 10／6 | cues | 8／8 |  |  | OAP211 18／6 |
| 2 P | 23／8 | OF33 | $3 / 8$ | 12ay | 16／6 | 4： 51 | 143 6／－ | Ecrixy | $11 / 6$ | F2ms | 4／－ | PLL＊ | $1 / 8$ | vexess | 8／3 | ${ }^{4} 42$ | 2015 | OCl6W 35／－ |
| $\bigcirc \times$ | 3 3 |  | $2 / 6$ | İA4\％${ }^{\text {d }}$ | 8／1 | $4{ }^{4} 101-$ | $1177 \quad$ 8／8 | Ecexad | $18 / 4$ | E：\％ |  | PCLas | $7 / 8$ | lurso | 8／6 | Wes | $24 / 8$ | $4 \geq 28 /-$ |
| 3.44 | $3 / 0$ | 6 Ht | 1／6 | 12abi | 9／L | 4 4i76 15／－ | 1）．4．3： 719 | Ecch mot | 15／－ | $\mathrm{Fr}_{4}$ |  | P＇list | \％－ | Ler121 | $18 / \mathrm{C}$ | 1 lis | 10／6 | 0623 57\％ |
| $3 \mathrm{~A}, 5$ | $6 / 9$ | 6J56 | 3／－ |  | 81 － | NAF $21 / 10$ | bham 3／6 | ECHE | 6／－ | in 13 | 14／6 | P1－85 | 2／8 | CTH4 | 7／－ | $\mathrm{H}_{76}$ | 9／6 | W＊5 1E／－ |
| $3 \mathrm{B7}$ | $5 /-$ | ${ }^{\text {bS5O＇}}$ | 4／3 | 12A 47 | 8／－ |  | 11．A／96 5／9 | WCrna | 618 | FC134 | 17／－ | P（Latb | 819 | THCHE | $18 / 6$ | W：7 | $2 / 8$ | （1\％$\%$ 25／－ |
| 31.0 | ． $3 / 8$ | cido | 3／－ | 12aty | 8／－ | 5uca 8／6 | D¢980 6 | Et5xti | 11／6 | FH4 | 118／6 | P1 Lam | 12／6 | ELLs： | $7 / 8$ | प HIM | $5 / 9$ | 0028 14／6 |
| 344 | －5／8 | ${ }^{\text {fiJ7 }}$（ | 4／6 | 12n＇4 | 4／6 | sth－3M6i40／8 | 104 12／8 | PFros | 4 24／－ | FW4／ | 10／6 | PINaL |  | U183 | $8 / 8$ | W 101 | $20 / 2$ |  |
| yusit | 7－ | ¢JJcas | $71-$ | 12AF | 3／4 | 51 Lethit 688 | $115410 / 8$ | ECH3 | 23／3 | C．TH0 | 9／9 |  | 25／－ | 1＇41 | $0 / 9$ | ${ }^{W} 107$ | $10 / 6$ | UC35 18／－ |
| ： | 4／6 | ¢5 6 | 12／8 | 12A16 | 5／8 | 0！ht 14／6 | 115T4 8／－ | E1H21 | 9／9 | clide | 52／8 | PEN20 |  | 1F＋ | $9 / 9$ | W749 | $17 / 6$ | Oc36 21／8 |
| 344 | $5 / 3$ | thod | 516 | 12A17 | 4／t | 53ht 14／6 | 1）6＂15 $\quad$ \％ 6 | ECH33 | $22 / 8$ | （1\％：30 | 71 | HEN40 | 11 | 1 l＇su | $8 / 8$ | 114 | 710 |  |
| 411 | $3 / 8$ | 4157： | 1／3 | lizavo | $6 /$ | 7： $0 / 6$ | 1）F33 8／8 | E1－Hxs | 0／8 | ¢\％：3\％ | $7 / 6$ |  | 34／－ | tFbs | 6／6 | ${ }^{\chi 18}$ | $6 / 8$ | （4．42 ${ }^{8 /-}$ |
| ${ }^{514} 4$ | $8 / 8$ | ¢¢ 76 | 4／－ | L2AN＇ | 4／1 | Br b／ | 1） 1 bif 15／－ | EAH4： | 7／6 | 1283： |  |  |  | CProb | $1 /$ | $\times 24$ | 18／6 | $\begin{array}{ll}0483 & 12 / 8\end{array}$ |
| 514 | 7\％ | \＄640 | $3 / 8$ | 12417 | 9／9 | im 4／8 | 11173 30／ | ECHM］ | 5／9 | 123 | 101． | PENS |  | Lfoss | $61 /$ | X41 | 151－ | $0^{0} 44^{8 / 8}$ |
| 00： | 4／6 | ¢N80． | \％ 18 | 12lsab | 5／9 | 3 | 112\％2／3 | EAH53 | 6／6 | 1：23 | 14／8 |  | 12／－ | l＇Lal | $8 / 4$ | X 01 |  | OC＋4P思11／－ |
| 5 CH | $7 / 6$ | Un：${ }^{\text {co }}$ | 241－ | 12135id | 4／4 | $\because \quad 22 / 6$ | Dr＊i 5／9 | Elhas | 9／6 | 1t：31 | $51-$ | tri | C18 | 11.44 | 28／8 | X 43 | $8 / 8$ | Oc．45 8／－ |
| SY8u | $4 / 3$ | till 1 | 101． | $1 \because \mathrm{BH}$ | 8／9 | 334 8／－ | 11497 10／－ | EClam | $6{ }^{6}$ | $\mathrm{H}_{13}$ |  | －ENSASt | ：110／3 | tiadi | 376 | xu4 | 4／6 | OC45PM10\％ |
| 3 Y 4 | $9 / 6$ | bildi | 6／6 | 1．E1 | 18／2 | miaz 6／6 | 1） H30 $^{15 / 4}$ | ECLC： | 71 | HABC | 800／8 | 1rendis | 410 | L． N | 61. | 30， | $6 / 8$ | U＇65 29／8 |
| 0\％3 | $71-$ | 6． 6 m | $81 /$ | Whatir | $1{ }^{1 / 6}$ | gilali liz／6 | 114633 4／8 | E1 Las | 8／6 | HLP | 7／6 |  | 10／－ | 1314 | 15／2 | $\mathrm{X}_{5} \mathrm{it}$ | 218 | U＇iss 26\％ |
| $5 \mathrm{K4}$ | $7 \%$ | b17．7：T | $4 / 6$ | 12．54．7 | $2 / 6$ | 671AV 67／6 | 1）1785 3／8 | kc Lat | $8 / 9$ | HL13C | 4／－ | 16SA4 | 7． | 1 My | 6／10 | $\times 18$ | 9／－ | 0170 d／6 |
| 6／301 | $8 / 3$ | ${ }^{11.17}$ | $12 / 6$ | 1：Jこ：T | 718 | wnce 421－ | 1167 3／9 | bel | 2076 | 1112； | 11／6 | A $\mathrm{B}_{4}$ |  | 1 1200 | 812 |  | 8018 | 0671 3／6 |
| 6．${ }_{6}$ | 8／9 | bl」k | 101－ |  | 10\％ | घuc＇42／－ | 114ns 23／3 | E\％ | $20 / 6$ | H23\％1 | 1）5／－ |  | 28／11 | Cric | $6_{6}^{6 / 4}$ | $\times 7$ | $27 /-$ | O6\％8／－ |
| ${ }^{\text {didandi }}$ | 5／9 | 61．193 | 918 | 12Kこせ？ | I $3 /$ | 906＇t 16／－ | DH101 201－ | $\mathrm{Hr}^{2}$ | $8 / 6$ | 1121 | 319 | PES／H1 |  | ！co | \％／－ | X 814 | $29 / 1$ | 0773 18\％ |
| ${ }^{8} 4158$ | 4／－ | BLis3 | $8 / 6$ | Hhmel |  | 15042 18／0 | DH14716／11 | ER゙3t | $3 / 8$ | H1．41 | 18／6 | 4021 | $17 / 6$ | $\cdots$ | 9／－ | ${ }^{\text {d }} 101$ | 29／6 | Oit 81－ |
| O41\％ | $3 /-$ | 11．1933 | 31 |  | 3／8 | 1502\％4／8 | 15632 79 | Hr：34． | 1／6 | 111．2211 | 1）8／6 | P1，3 | $8 /-$ | Cも | 7／6 | $\times 108$ | 28／－ | 0078 8／－ |
| bacs | $2 / 8$ |  | 3／8 | Lamat | 819 | $1318181 /$ | 1 k du 15／6 | Ef 34 | $8 / 9$ | 1 |  | P1，36 | $8_{1}{ }^{\text {d }}$ | 1＇8 | $11 / 6$ | ＋114 | $9 / 8$ | crinis 8／6 |
| $\underline{5497}$ | 5／9 | ＋N76 | $5 /$ | 1284 7 | 4／－ | 186 B ＇］34／11 | 1J69 4／－ | Er 40 | $8 / 5$ |  | $4 / 6$ | 1．1．3n | 18／－ | L＇t | $5 / 3$ | X 119 | $8 / 6$ | O67 12／－ |
| ciAls | $8 / 8$ | $\mathrm{BiP}^{1}$ | $9 / 3$ | 12mai | \％／ | $\because 158016$ |  | EFP1 | 6／8 | His304 | $28 /$ | PLxI | $8 / 9$ | 11\％ | 4／8． | X 142 | \％1－ | $00^{6} 81-$ |
| 0.4 K 5 | $4 / 9$ | $\mathrm{OF}^{125}$ | $6 / 8$ | 128H7 | $3 /-$ | 2－2ks $10 / 6$ |  | EFt2 | $4 / 8$ | 1才年 | $8 / 3$ | P1起 | $3 / 3$ | Ly | 1019 | Y 03 | $5 /-$ | $0 \times 1$ 4／－ |
| OAtis | 12／8 |  | 9／－ | 1285\％ | $61-$ | $301201-$ |  | EFsu | 816 | HVR2A | 819 | 1PR： | $81 / 3$ | $1 \cdot \mathrm{y}$ | $8 /-$ | ${ }^{6}$ | 5 5－1 | Oris $4 / \mathrm{F}$ |
| 6AKM | 5／8 |  | 11／8 | 12\％ 6 | 3／－ | 30： $10 / 6$ | 113．3 6／8 | EFS | 8／－ | 1W： | $5 / 8$ | Pl，N4 | 5／6 | LY＋1 | $5 / 2$ | 263 | 5／6 | O484m 81 |
| －iALA | $2 / 3$ | By7\％ | 1／4 | 12 m 67 | $81 /$ | 303 15j－ | 1）113 5／6 | 1．178 | $51 /$ | $1 \mathrm{C} / 1350$ | 11516 | Plutur | $18 / 8$ | 1 $\mathrm{Y} \times$ | $5 /-$ | － | $7 / 8$ | OR83 101－ |
| iAMS | $2 / 6$ | n¢75］ | 71 | 124．87 | $5 /-$ | 04 151－ | 11160 15／－ | 1．1．54 | $3 / 5$ | 14：\％ | 118／5 | Prat | 43 | 1＇20 | $1-$ | 277 | $3 / \mathrm{H}$ |  |
| ${ }^{\text {JAMG }}$ | $3 / 9$ | tikiti | 5／9 | 12lum | \％ | 13／－ | 1172 15／－ | Eras | $9 / 9$ |  | $20 / 6$ | $\mathrm{Pr}^{15}$ | 10／－ | c19\％］ | 718 | 7729 | ${ }_{6}^{6 /-}$ | OLisd 8／－ |
| 9A4 | 5／9． |  | 11\％ | 1以34 | $2 /-$ | ：106 13／－ | 11.75 301－ | EFros | $91 /$ | $\mathrm{KF}^{2} 36$ | 12／6 | PX | $9 /-$ | $1 \cdot 16$ | $15 /-$ | 7\％ 79 | ${ }^{8 / 3}$ | 0¢13 $12 \%$ |
| ${ }^{18} 4.15$ | 201. |  | 419 | ${ }^{13131}$ | $5 /-$ | Hitsa $12 / 6$ | 111．42 ${ }^{1 / 6 / 6}$ | 1\％к产 | 81. | ${ }_{\mathrm{K}}^{6} \mathrm{~L} 3 \mathrm{3}$ | 11／6 | $1 \times 25$ | ${ }_{8 / 1}^{8 / 8}$ | 115 | 5／6 | 2759 | 38／－ |  |
| ©A＇t | $3 / 9$ | isc＇7 | $4 / 8$ | 15113 | 5／6 | （162＇） $12 / 6$ | 11.943 | Wx\％ | 4／－ | K1．L | 21／4 | $1{ }^{1} 131$ | $61-$ | 1118120 | 816 | Tranniz | turs | 0¢170 8／8 |
| ${ }^{\text {sictio }}$ | $3 / 3$ | 6n937 | $4 / 9$ | 14B4 | $20 / 8$ | 403s $15 \%-$ | 11290 6／－ | LFPat | $3 / 8$ | ¢T2 | 5\％－ | 1 | $8 / 8$. | ¢19 | 48／6 | and diod |  |  |
| 6.4 | 5／8 | ¢nH7 | $3 /-$ | $1+117$ | 818 | Hfin7 71／－ | d1／itio 5／8 | L．FH2 | 2／6 | KTM | 15／－ | 11：38 | $3 / 5$ | 122 | $5 / 9$ <br> $19 / 8$ | AA12y | 14／6 | O2\％ |
| ${ }^{1177}$ | 5／8 | 1is 57 | $4 / 6$ | 1， | 10／6 | 8103 718 | DIstO 10／6 | ${ }_{\text {Preva }}$ | 11／8 |  | 5／6 |  | $5 / 5$ |  | 12／6 | Aclin Acder | 14／6 918 |  |
| $1188 G$ 6 BAU | 2／8 | ${ }_{6}^{\text {ti8k }}$ | 4／8 | 198： | $6 /-$ $12 / 6$ | $\begin{array}{ll}7193 & 1 / 6 \\ 7475 & 2 / 9\end{array}$ |  | Erys | ${ }_{7 / 9}^{101}$ | K＇1330 $\mathrm{K}^{\prime} 26$ | 24／1 | ${ }^{\text {PY }} \times 1$ | 5／－19 | （1－5 | $8 / 6$ $7 / 8$ | ACle | 9／8 25／－ | $\begin{array}{ll}\text { Oxpl } \\ \text { OCPI } & 14 /- \\ \text { 17／6 }\end{array}$ |
| 6BE6 | $4 / 9$ | tisN7 | 41. | 14 | 10／6 | A1834 20／－ | 1）W4／3508／6 | EFlud | 719 | KT41 | 7／0 | PY83 | $3 / 8$ | 1－1 | 819 | AF10： | 27／6 | ORI＇12 12／8 |
| ${ }_{6} \mathbf{H C i t i g}$ | 13／6 |  | $5 /$ | $14 \mathrm{AQ5}$ | $7 / 3$ | ACOLE 23／8 | nW $+/ 5008 / 6$ | 1：1190 | 71 | KT44 | $5 /-$ | PY＇ | $7 / 8$ | C：33 | 13／6 | AF114 | 11／＊ | 8xtil 10\％ |
| a Bhes | $5 / 3$ | 6RR7 | 12／6 | 10BCEGG | 20／5 | AcO44 8／－ | DY86 7／t | EK！ 2 | 25／11 | $\mathrm{h}^{\mathrm{T} T 1}$ | $8 / 8$ | Prever | 6／6 | U35 | $16 / 8$ | AF115 | 10／6 | T82 12／6 |
| 0BJ6 | 5／6 | 6887 | 2／－ | $19 \mathrm{H1}$ | $81-$ | AC2HL 1016 | DY47 8／－ | EK32 | 8／8 | $\mathrm{K}^{\text {Tu3 }}$ | $3 / 9$ | PYM01 | 71 | ${ }^{1} 37$ | 291－ | AF116 | 10\％－ | Tx8 15／－ |
| ${ }^{6} \mathrm{BLQ5}$ | 4／8 | 61J40T | 8／8 | ＊411 | 101－ | Ac＂pen 11／6 | E801：94／－ | ELL | $19 / 6$ | KTVR | 12／3 | QP？ | 5／－ | U44 | ${ }^{5 / 8}$ | AF17 | 518 | F10／15A12／－ |
| fBu7a | $7 / 6$ | 0 CJO | 51. | 20 L 2 | 21／－ | A C＂z2＇EN／ | Ex3F $24 /$－ | ELL32 | $3 / 8$ | ${ }^{1} \mathrm{~T}$ T4 | 12／6 | $\mathrm{QPP}^{\text {P202 }}$ | $12 / 6$ | C＇45 | $15 / 6$ | AFH18 | ${ }^{801}$ | XA102 19／6 |
| ${ }^{0687}$ | 8／3 | bU | $7 /$ | 2012 | 11／6 | （1）12／6 | （C）10\％ | E143 | $8 / 9$ | KT88 | 28／－ | QP25 | 5／－ | 147 | $8 / 6$ | Ar124 | 11／－ | $\times \mathrm{AlO3} 15 /-$ |

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where NI is the current multiplication required on the lowest current range. In the writer's case the lowest range was 3 mA and the f.s.d. $660 \mu \mathrm{~A}$, giving $\mathrm{Nl}=3.000 / 647$, and Rm was 23.3』. So, in this particular case,

$$
\begin{aligned}
& \mathrm{R} 1+\mathrm{R} 2+\mathrm{R} 3=23 \cdot 3 /\left(\frac{3.000}{647}-1\right)=6 \cdot 4 \Omega \\
& \text { Step 2. } \mathrm{R} 2+\mathrm{R} 3=\frac{\mathrm{Rm}+(\mathrm{R} 1+\mathrm{R} 2+\mathrm{R} 3)}{\mathrm{N} 2}
\end{aligned}
$$

For example, to furn the w.r...r's meter into a 30 mA meter,

$$
\mathrm{R} 2+\mathrm{R} 3=\frac{23 \cdot 3+6 \cdot 4}{30 /(0 \cdot 647)}=0.66 \Omega
$$

It follows that $\mathrm{R} 1=6 \cdot 4-0 \cdot 66=5 \cdot 74 \Omega$.

$$
R m+(R 1+R 2+R 3)
$$

$$
\text { Step 3. } \quad \begin{aligned}
R 3 & =\frac{N 3}{23 \cdot 3+6 \cdot 4} \\
& =\frac{200 /(0 \cdot 647)}{300}=0.066 \Omega
\end{aligned}
$$

for 300 mA in the case in point. Thus $R 2=0.66-0 \cdot 066=0 \cdot 594 \Omega$. Clearly the method cian he extended to any number of current ranges. Thus a 3 A range would require an extra resistance R4 of $0.0066 \Omega$ ) and so on.

## Making the Shunts

The largest shunt resistance $(5.74 \Omega)$ lies in the range of resistance which can be measured reasonably accurately by the original unit. However. the writer decided to try his luck without actually measuring the resistance but instead to cut suitable lengths of wire of known resistance. In fact the only wire available was ordinary copper wire. but this proved to be quite satisfactory. (Resistance, values are given in standard wire tables.)

It is necessary to use a reasonably long piece of wire, because one can never say within a quarter of an inch or so what the final length is, because a certain amount of wire is in effect lost in the connections. (It is reasonable to allow an extra half-inch.) If the length is short, say, a couple of inches, this uncertainty accounts for rather too high a proportion of the total. On the other hand. it is tedious to wind huge lengthe of wire. A search in the wire box revealed some 43 s.w.g. $(2 \cdot 359 \Omega / y d)$ which would serve for the $5 \cdot 74 \Omega$ resistor ( 87.5 in .) and some $36 \mathrm{~s} . \mathrm{w.g.g}(0.594 \Omega / \mathrm{yd})$ which would do for the 0.662 resistor ( 38.4 in .). These resistors were made up and connected in series across the meter. If correct the meter must read very nearly 3 mA . since the third resistor ( $0.066 \Omega$ ) has little effect on the 3 mA range.

A current of 3 mA was obtained from a 3 V dry battery and a $1 \mathrm{k} \Omega$ resistor (later to serve as the 3 V " multiplier") and the meter registered very nearly f.s.d.. showing that the shunts were correct. No suitable single wire was available for the third shunt so a piece of $7 / 0.0076$ (i.e. seven strands of 36 s.w.g.) connecting wire was tried. The Iength, calculated on the assumption that the resistance was one-seventh of the resistance of $36 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. was $31 \cdot 2 \mathrm{in}$. It turned out that the meter read $25^{\circ} \%$ high in the 300 m A range. This seemed odd until it wat realiaed that to use stataded wire one has to allow for the twit?

The length was adjusted. using a 30 mA standard current ( 3 V . $100 \Omega$ ) so that the meter read f.s.d. on the 30 m A range and one-tenth f.s.d. on the 300 mA range.

## Voltage Ranges

By leaving the 3 mA shunt permanently in circuit a range of 3 V was obtained using $1 \mathrm{k} \Omega, 30 \mathrm{~V}$ with $10 \mathrm{k}!$. and 300 V with 100 ks . Close tolerance resistors should be used $( \pm 1 \%)$ and the power rating should be adequate. Thus at f.s.d. the $100 \mathrm{k} \Omega$ resistor must dissipate $3 \mathrm{~mA} \times 300 \mathrm{~V}=$ 900 mW . so a lW resistor is required.

In addition a $300 \mathrm{mV}(0.3 \mathrm{~V})$ range was included :


Fig. 8: The conventional ohmmeter which with a non-linear scale produces inaccuracies at the low-resistance end.
this is useful for work with transistors. On a low range like this, the meter resistance ( $5 \Omega$ with the 3 mA shunt) becomes significant. However, a $100 \pm 1 \%$ resistor was used as it was the nearest standard value and the $5 \%$ error was felt to be tolerable on this range.

## Resistance Ranges

An ordinary ohmmeter (Fig. 8) has a non-linear scale and is only accurate for resistor values which produce about half-scale deflection. Since the surplus unit had a good calibrated 50S resistor in it (actually calibrated $0-25 \Omega$ ) and since the meter was centre-zero it was decided to use a


Fig. 9: A three-range bridge circuit for resistance measurements.
bridge circuit for resistance measurements.
The original instrument was a single-range one, but it is possible to make use of the calibrated resistor in a multirange instrument. To do this the circuit has to be rearranged as in Fig. 9. Heıe the 100 ? ? nominal resistor should be exactly double the value of the $50 \Omega 2$ nominal variable resistor. To make sure that this was so use was made of some of the components in the original circuit.

Referring to lig. 4 it will be eeen that the right-


Fig. 10i The complete multimeter circuit.
hand ratio arms are in the ratio 2:1. The technique is to use them as standards to select a suitable $100 \Omega$ nominal resistor. To do this the positions of the $50 \Omega$ variable and the $X$ terminals are exchanged. The $50 \Omega$ resistor is set to maximum and left alone. Then various resistors of $100 \Omega$ nominal value are tried until one is found which balances the bridge. This resistor has twice the value of the variable resistor and is used in the final circuit.

The resistance standards (Fig. 9) should be closetolerance components. RS is a safety resistor to


Fig. II : Illustrating a suitable handie for the meter which may double as an inclined support.
protect the meter against overload. A study of Fig. 9 will show that the meter receives the largest current when on the $25 \Omega$ range with nothing connected to the $X$ terminals and with the $50 \Omega$ variable resistor set to zero. Then, if there were no RS, only the $50 \Omega$ standard would be effectively in the meter circuit. RS is chosen to limit the current to a safe value under these conditions.

It was originally intended to use the meter with its 3 mA shunt for the bridge, but the sensitivity was poor on the $2,500 \Omega$ range. So in the final circuit it was arranged to disconnect all the meter shunts when measuring resistance. This meant having an underdamped meter but it takes a few seconds to find the null anyway, so the disadvantage is not so great.

## ON THE SHORT WAVES

-continued from page 240

An undisclosed 25 m band frequency has replaced $15.265 \mathrm{kc} / \mathrm{s}$ for the transmission from 1830-1925 of the Radio Voice of the Gospel, P.O.B. 654. Addis Ababa. Ethiopia.

There are new frequencies for all the English transmissions to Europe of Radio Warsaw, Warsaw, Poland. The new schedule is 1830 1857, 1930-2000 and 2130-2155 on 7125/7, $285 \mathrm{kc} / \mathrm{s}, 2030-2100$ on $6,135 / 7,270 \mathrm{kc} / \mathrm{s}, 2230-$ 2300 on $7,145 / 7,270 \mathrm{kc} / \mathrm{s}$ and $1,502 \mathrm{kc} / \mathrm{s}$ medium wave. There are also multilingual music transmissions from $1900-2030$ on $7,145 / 1,502 \mathrm{kc} / \mathrm{s}$ and 2330-0100 on $9,675 / 7,125 / 1,502 \mathrm{kc} / \mathrm{s}$.

Cairo Radio's transmission to East and Central Africa from $1545-1830$ is now on $17,920 \mathrm{kc} / \mathrm{s}$, where it is giving good U.K. reception. English is at 1745. This station wants reports, which should be sent to Propagation and Monitoring Department, U.A.R. Broadcasting and TV, Maspero Cairo, Egypt.

Finally N.H.K., Tokyo, Japan, now uses 11.780 / $15.135 / 17.870 \mathrm{kc} / \mathrm{s}$ for its European transmission from 0615-0845. The English portion is from 0800-0845.

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| G4HZ writes $\therefore . \quad \mathrm{mm}$ delighted with it, it improves my Eddystone 640 in all respects. IThe difference with the Preselector is fantastic, a weak I signal on 15 metres about S 2 changed to S8. On the L.F. Bands, unwanted noise and I mush is cut out.
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## SINGLE SIDEBAND RECEPTION

SIR,-I wonder if users of simple short wave receivers have noticed how well single sideband transmissions can be tuned in with these sets. A test, and a little practice in tuning such s.s.b. signals, is best first made on the 80 m amateur band. With an ordinary superhet, or with a t.r.f. receiver adjusted so that the detector is not quite on the point of oscillation, any s.s.b. signals will be quite unintelligible. If regeneration is then advanced very slightly. so that the detector is just oscillating, the s.s.b. can be recejved very well. Receiver tuning is adjusted so that the detector frequency is near the frequency of the eliminated carrier of the s.s.b. transmission. On the $14 \mathrm{Mc} / \mathrm{s}$ band, many USA and other amateurs will be heard, using this system.F. G. Rayer (Upton-on-Severn, Worcestershire).

## IF YOU CAN'T BEAT 'EM-JOIN 'EM

SIR, With reference to the Leader Competition entry from Mrs. Joyce Burdett (page 62, May issue) who claims to be a "wircless widow". Why should she feel like this"? The old adage, "If you can't beat 'em, join 'em" is the best cure for the XYL's blues. There are now well over thirty YL'S and XYL's in this country with their own callsigns and I don't think there is one amongst us who could be described as a blue-stockinged technical type; a R.A.E. pass does not require brains, only a little application of effort and is within the reach of all.

With her own ticket Mis. Burdett would be in a position much envied by the unmarried licensed YL's and those XYL's who have their own ticket but whose OM's have no knowledge of radio to help them. She would have her (or even his) equipment built and maintained and her aerials rigged for her, unlike some XYL's with licenses who have to do their own construction and maintenance, and are even called in by the OM to mend a fuse!

Many radio societies, like our own, welcome YL's and XYL's and even the OM's of amateurs as members, though there are as yet, a few societies who have not seen the light and condescend to welcome their ladies only to the annual dinner as decoration (or more correctly, to ensure that the OM's do not get into trouble for being late home!)

How would your correspondent feel if she were the OM of a lady amateur-with strange men calling on her as the result of a mobile QSO, various men collecting her at oud hours to do a stint of Field Day operating or all night on a Scout Jamboree station?

I am sure that if some OM's can be tolerant of their wives' hobbies, whether they be knitting,

Whilst we are always pleased to assist readers with their technical difficulties, we regret that we are unable to supply diagrams or provide instructions for modifying commercial or surplus equipment. We cannot supply alternative details for meceivers described in these pages. WE CANNOT UNDERTAKE TO ANSWER QUERIES OVER THE. TELE. PHONE. If a postal reply is required a stamped and addressed envelope must be enclosed with the coupen from page iii of the cover.

The Editor does not necessarily agree with the obinions expressed by his correspondents
bingo or amateur radio, then it is up to the XYLs to do the same. I joined my husband's hobby (not radio) because 1 couldn't hope to beat it and he now even tolerates mine! After all, amatelur radio is a hobby which can, to a large degree, be pursued at home, as long as the XYL doesn't drive the OM to the last resort of the henpecked hamnamely, mobile operation on his way to work! "G3XYL" (name and address supplied).

## MODIFICATIONS FOR P.W. SIXTEEN MULTIMETER

SIR.-I have recently completed the P.W. 16 Multimeter (deseribed in your January 1964 issue) including modifications to extend the ranges available to eighteen. I am very pleased with the results achieved, and the general professional appearance of the completed instrument.

Construction of the instrument was straightforward, provided carc was taken and the job not rushed. This applies particularly to the installation of switch $\mathrm{S}_{2}$. which is PVC insulated and very difficult to solder without damage. However, this was successfully accomplished by using a heat shunt and by not keeping the iron in contact after the solder had run.

The two additional ranges included in my meter extend the voltage readings up to $1,000 \mathrm{~V}$ a.c./d.c. (Messrs Lasky Radio supplied the additional components required for seven shillings.) Two additional terminal sockets were fitted by drilling the escutcheon plate in line with existing holes to be found in the left-hand comer of the meter case when viewed from the rear. These two sackets then became the common positive and negative terminals of the published design. The original sockets of the meter are then used for the 1.000 V ranges.

The positive socket becomes the 1.000 V d.c. positive, and the negative the 1.000 V a.c. The new common negative socket is used for all ranges.

Two additional $5 \mathrm{~m} \Omega$ resistors are wired in series between the $1,000 \mathrm{~V}$ d.c. positive socket and terminal 5 of SI and two $250 \mathrm{k} \Omega$ resistors are wired in series between terminal 5 of S1B and the 1.000 V a.c. socket. These additional resistors are housed vertically either side of the insulated pests used for holding the case together The double solder tag " B " is not required and the connections originally taken to this point are soldered directly to the common negative terminal.

These modifications, besides making the meter more versatile. will enable use to be made of the Taylor HV probe which extends the range up to $25 k$ V.-L. J. Lyon (Lexden, Colchester, Essex).

## TRIMMING TOOL OR TUNING TOOL

SIR,-I wish to thank you for the really valuable
Trimming Tool Set which accompanied the April P.W. It will obviously be most useful for its intended purpose, but I have found that in an emergency, $l$ can also use the No. 4 tool as a tuning-fork-as its shape, but not its material might suggest.

As a violin player, I always carry in my case a $440 \mathrm{c} / \mathrm{s}$ tuning fork, but I have now found that if I grip the No 4 tool by the "plug" and gently hold the tines of the fork between finger and thumb of my other hand and then slide these digits off the fork, a tone equal to A-flat; i.e., just a semi-tone below the now standard pitch for "A" (Anatural); is emitted.

The sound is very quiet, of course, and I have to hold the tool close to my ear whilst stroking the tines, to hear the note. - L. F. HUNTER (Rugby, Warwickshire).

## A CLOSED SHOP?

SIR.-I feel that I must object to one point in H. Naylor Caley"s letter in May P.W. Although his suggestions for further study at evening classes are sound, I fear he gives the impression that all radio clubs are kinds of secret societies in the mystic of radio, and are restricted entirely to "ham" radio.

I would point out that most radio clubs are strictly "amateur radio societies", and although one of the main interests is usually ham radio, you will find members with interests in all fields of radio and electronics.

I joined my local club (Eccles and District) only six months ago, but in that time I have received encouragement and assistance in building an 11valve double conversion receiver, more than doubled my technical knowledge, and become a member of the R.S.G.B. Hardly an example of a closed shop of licensed amateurs with very restricted interests. - W. R. Bannerman (Flixton, Manchester).

## P.W. GUITAR AMPLIFIER

S
IR,-I was rather puzzled to read in the article on the 35W Guitar Amplifier (P.W., May 1964), that many thousands of volts are present at the top cap of the 807 valves when the signals are present.

High anode voltages may he expected with inductive anode loads, (cf. line output transformers), but with an output transformer correctly loaded, the transformer behaves as an impedance matching device and not as an inductance. The anode load is therefore resistive and is equal to the transformer secondary load reflected back to the primary.

807 valves commonly exhibit internal fluorescence, which could give the impression of very high anode voltages when only relatively low voltages are present.

If the secondary load on the output transformer were removed, then it would not only be possible to break down the transformer insulation due to the generation of high potentials by what would be now an inductive load. but also cause risk of damage to the valve. There is the possibility of flash-over in the valve due to high voltages but
also the possibility of increased screen dissipation due to a lowering of anode voltage.-F. W. Adams, B.Sc. (Sheffield, 10).

SIR,-With reference to Mr. Adams' letter, I would like to clarify the paragraph referred to in my article.

To recapitulate, the sentence under discussion reads: ". . . keep away from the 807 top caps, high peak voltages of many thousands are present here when a signal is present, and severe burns can be had if they are accidentally touched during use".

What is meant by this is that, should the caps be touched (especially if the other hand or a part of the body is in contact with the chassis) a spark will be drawn between the top cap and the person's hand. Due to the spark comprising an alternating current component, the back e.m.f. across the output transformer primary winding is extremely high. The actual voltage developed depends upon the spark frequency, and the level of signal present in the amplifier at the time.

Placing part of the body across this circuit point, puts a relatively low resistance across the transformer, which of course upsets its matching. Once contact has been made and a spark drawn as the unfortunate constructor endeavours to move away, the transformer behaves as an inductance purely. and as stated, due to the increasing back e.m.f. produces a vicious circle with the spark voltage increasing until the gap is too large or the external resistance of the person's body has increased to a point where the current flow is too low to sustain the spark. I have had first hand experience of this effect, and nursed a nasty burn on one thumb for quite a few weeks afterwards.

Incidentally. Mr. Adams is quite right in his remarks-"high voltages are not present whilst the amplifier is operating normally", the anode voltages of the 807 's being 450 V under all con-ditions.-Brian L. Phillips (Preston, Lancashire).

## Sir-I would be grateful if any reader could sell or loan me...

iscues of P.W..-P The February, March and September 1962 issues of P.W.-P. Cunningham, 20 Kenwood Gardens, Gants Hill, Ilford, Essex.
. details about the Ekco Car Radio, No. CR74, especially of the external connections to the sockets.-P. Bayliss, 18 Stuart Road, East Barnet, Hertfordshire.
the February, March and April 1962 issues of P.W.-B. J. Collins, 54 White-Lion Road, Amersham, Buckinghamshire.
SYMONDS, $66 \dot{\mathrm{St}}$. Barnabas Road, Cambridge.
the information regarding the R1155: power, output, etc. All data received returned within 48 hours.-T. L. SadLer, 10 Sheaf Avenue, Abbeydale Road, Sheffield 7.
larly the Mk. 3 No. 2 supply unit. Mk. AUBREy, 1! Addersley Road, Norton, S.o.T., N. Staffordshire.
. the service manual for the Sobell RS402 a.m./f.m. receiver.-F. CLARKE, 57 Woodcroft Road, Smitha.m./f.m. receiver.-F. Clarke, 57
down Road, Sefton Park, Liverpool 15 .
.. circuit, valve line-up and any information on ex-government type No. MCR1,-R. Davis, il Bishop's Lane, Robertsbridge, Sussex.
"R . . . written or printed notes leading to the MOT "Radar-Maintenance Certificate" standard or any other material on the subject.-M. FaHy, 1 Castle Street, Limerick, Eire.
$\ldots$ the circuit and any other data concerning the R 107 set ZA 3050 .-R. WILD, 140 Nansen Road, Alum Rock, Birmingham 8.
. . the September 1956 issue of P.W.-B. Dodos, 42c Alma Place, North Shields. Northumberland.
circuit and/or manual of the Romac Radio Corporation Television set. Model 179/T.-J. McGrail, 494 Rochester Way, Sidcup, Kent.

No. 3
JUNE/JULY 1964

## ANOTHER ACCESSORY FOR THE MICRO-6

Because of the enormous interest in the Micro-6, we at Sinclair Radionics have been devoting much of our time to the development of accessories which add to its usefulness. The first of these, the TR 750 power amplifier was introduced last month. This remarkably low priced high performance design has, of course, a great many applications but in conjunction with the Micro-6 or the Slimline it can form a really powerful car, home or portable radio.

The latest Micro-6 accessory is just as exciting! We have designed and produced an elegant nylon wrist strap which clips straight on to the Micro-6 converting it into a wrist radio-the first of its kind in the world! The earpiece lead may then be run up the sleeve and the radio listened to without any sign orrailing wires.

The wrist strap may be removed instantly when desired, but in use it is firmly attached and the set cannot come loose. It will add immensely to the pleasure and usefulness of your Micro-6, particularly when out of doors.

## THE MICRO-6 CIRCUTT

Many readers may be curious to know how we achieve the high performance we claim for the Micro-6 in a set of such tiny size. That our claims are genuine is, of course, proved by the countless constructors' letters we receive, more of which we give below, but you may like to know more about the design. The circuit is a completely new one based on our micro-alloy transistors (MATs). These tran-

## FROM OUR POST BAG

Enthusiastic letters continue to reach us unabated. A constructor from Aylesbury writes:-
"Many thanks for having dispatched my order of your little receiver the Micro-6. I built the set on Saturday evening and in less than two hours; the results being obtained straightaway. Home, Light, and Third, received on more than ample strength, and at least 12 foreign stations, such as Luxembourg, Paris, two German stations, and others. all these at more strength than needed, with good selectivity. Clarity is very good; and no fading on Luxembourg. Truly a remarkable achievement, to be heard and seen, indeed to marvel at less than a Match Box," (Signed) J.G.
D.M.F., Southport, states:-
"As one who remembers the days of bright-emitter valves and has seen radio grow from that to colour television I can still marvel at the design and performance of your anlazing little Micro-6. Having built it, I find it hard to be without. I seem always to have use for it, and I find it abwolutely fascinating that so much entertainment can be had from so tiny a set."
(Signed) D.M.F.
sistors have higher gains than any other types at low voltage levels and, as they are also very small, are ideal for miniaturised equipment. The Micro-6 has two high gatin stages of R.F. amplification, a double diode detector and three stages of A.F. amplification. Thus we have used exactly the same number of stages of gain as are employed in conventional 6 transistor superhets.

We are able to do this because we use genuinely micro-miniature components not normally available to the home constructor. For example all the resistors are only $\frac{1^{\prime \prime}}{5}$ in length. These tiny components help to make the Micro-6 simple to construct because we don't have to pack them so closely together. If you would like more details just send $1 /$ - for our booklet on the Micro-6 which gives the complete circuit diagram and technical description.

## BRITISH BEATS THE WORLD!

We recently launched an export drive for all our products. We have already received enquiries for the Micro-6 from over 30 countries and samples are now going to agents throughout the world. One company in the U.S.A. has ordered 1,000 TR750 ansplifiers and 1,000 micro-amplifiers. We were particularly pleased by this result as the orders were obtained in the face of very tough Japanese competition.


This time our devoted enthusiast MIKE farad says

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| :--- |
| one for | Atahel, twin for thi hid and come tor ctramemina. The ing нill halic нill hatc

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THIS IS THE SPEC THAT GETS SUCH WONDERFUL RESULTS
The Micro-6 uses Micro-Alloy Transistors in a completely new eircuit. Two stages of R.F. amplification are followed by an efficient double diode detector which drives a highogain 6ostage A.F. amplifier. Powerful A G.C. applied to the first R.F. stage ensures fade-free reception from the most distant stations tuned in. Everything including ferrite-rod aerial and batteries is contained within the elegont tiny case.

SPORTS
EVENTS


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59/6
7/6


ACTUAL SIZE-2" $\times 2^{\prime \prime}$

## THE TR750 HAS TREMENDOUS POWER

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## BRILLIANT FEATURES OF

The performance characteristics are brilliant3'4 W. transformerless output with frequency response $L$ IdB from 30 to $\mathbf{2 0 , 0 0 0} \mathrm{c} / \mathrm{s}$-hi-fi by any standards. input10 mV into 2 K ohms for 750 miliwatts output for any standard 25 to 35 ohm loudspeaker. Operating requirements from 9 to 12 volts. Full building instructions are supplied with parts for the TR. 750 , which include latest type Micro-Alloy driver transistor, exciting new "Magnagain" output transistor.

## TR 750 POWER AMPLIFIER

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T．C．C．or Dubilier Tubular Condensers．
.5 tif $300 \%$ ．
.25 माf 300 \％．
.605 wt 800 v．
.61 mi 500 v ．
.0001 mil 1.000 v ．
.001 wif 1.0000 v．
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． 005 m Hf 1.000 v

10／－doz．
7／6 doz．
6／－dos．
6／－doz． 4／－duz． 8／－duz． 8／－thoz． 7／6 doz． 91－tしょa． $8 / 8$（luz．
$101-$ duz．

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0－50u microamp．thin．flush

20 Guctoatul．2fin．suriac
$17 / 6$
$27 / 6$
0.3 Enflamp． 24 m ．flush
$0-100$ militang． $2 \frac{1}{3}$ in．Aush
0.50 н milliaıup．2tia．Hush $17 / 6$
176 15／－ ．It ampr．，6／6． Noon Lamp－midget wire ended．Heal main tester，etc．2／－．Ex．Govt．1／6．
Phillips Trimmers－ $0.30 \mathrm{pF}, 1 /-$ ea． $\mathrm{B} / \mathrm{=} \mathrm{doz}$ ． Heavy Duty Test Prodg－red and black with plug－in lead attachwetits．6／6．
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Themporatiog ad］the thiremerta Anyone whe can xoldill wat make it．Many thentanals abreals in use all well the coustry， Full coverage of Long and Medium wave bands．
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