## PRAGTICAL Writeless <br> OCTOBER 1963

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FOR CURRENT, VOLTAGE, RESISTANCE AND WATTAGE

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PUSH-PULL AMPLIFIER $£ 5.5 .0$ (6/-Carr.)
Bramd new 200/240 A.C. mains. Hass, treble and yol. controls. With valves EZ80, ECC83 and 2.EL84 giving full 8 w. Chassig $12 \times 3 / \times 3$ in. With o.p. trans. for $2-3$ ohm speaker. Front panel (normally screwed to chassis) may be removed and usedas $\&$ Vol. Controls

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Type TR3. Fully buitt, high gain, low noise, printed circuit. Attractive grey and gold front pane $13 \quad x$ lifinHeight 5 lin. overall. Front to hack radio and ext. speaker lacks. Valves

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## TO FIND VOLTAGE

## $\mathbf{V}=\mathbf{I} \times \mathbf{R}$

Set kilohms at left-hand pointer, or ohms at right-hand pointer. Read volts above the milliamps scale.
Example: 10 mA is passing through a 2 kilohm resistor. Set left-hand pointer to 2 (kilohms) ; reading off 10 (milliamps) on Scale (A) gives the voltage drop as 20 V (Scale B).

## TO FIND CURRENT

$$
\mathrm{I}=\frac{\mathrm{V}}{\mathrm{~V}}
$$

Set kilohms at left-hand pointer, or ohms at right-hand pointer. Read current below volts.
Example: 30 volts is dropped across a 300 ohm resistor. Set right-hand pointer to 300 (ohms). Reading down from 30 (volts) on Scale B, the current is shown as 100 mA .

## TO FIND RESISTANCE

$$
\mathbf{R}=\frac{\mathbf{V}}{\mathbf{I}}
$$

Set volts above milliamps and read either kilohms at lefthand pointer or ohms at right-hand pointer.
Example: Current is 20 mA and voltage 60 V . Set Scale A so that the $20(\mathrm{~mA})$ is lined up with $60(\mathrm{~V})$ on Scale B. Reading left-hand pointer gives resistance as 3 kilohm.

## TO FIND POWER (Watts) $\mathbf{W}=\mathbf{I} \times \mathbf{V}$ or $\mathrm{I}^{2} \mathbf{R}$

Set right-hand pointer to volts (Scale B), when the wattage is read off above the milliamps on Scale A.

Example: Voltage applied is 250 V ; current drawn is 20 mA . Set right-hand pointer to 250 . Reading above 20 (milliamps) on Scale A, gives 5 (watts) on Scale B. To find milliwatts proceed as above but use the left-hand pointer.

## OHM'S LAW MEMORY AID



The relationship between voltage ( V ), current ( 1 ) and resistance $(R)$ is given at a glance in the left-hand triangle. Wattage (W), current and voltage are similarly indicated in the other triangle.
Place finger over the unknown factor and the formula is given
 by the relationship of the two uncovered characters.

## PRACTICAL WIRELESS

## Ohm's Law Calculator

for d. c. circuits
(Presented Free with
'Proctical Wireless'
October, 1963)

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$19 /$ Case for above，Fith 9 \＆sin．speaker，two－lode grey
Complete Eit．with Tape add Mierophooe．
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 PETT，less power，complete with \＆EF80 valies PYTE，with power，complete with 4 EH8O and 1 Ez80 MT2，with \＆i．2．8 1.2.
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udio chassia，atereo grata and push．pull output stages．

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Our latest completely portable translstor radio covering medium and long waves. incorporates pre-tagged circuil board. bin. heavy duty speaker. cop trade transiz condenser, wave change slide switch, sensitive fin. ferrite rod aerish, Push-pull outin. eut. Wonderful reception of B.B.C. Home and Light. 208, and B.B.C. Home Continental stations. Handsome leather look pocket size case, only $61 \times 3 t \times 1 \mathrm{n}$. approx.. with gilt speaker grille and supplied with hand and shoulder straps. Total cost of all 54.2 P. \& P. Parts Price List and pares only

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- 9 stages- 7 transistors and 2 diodes
Covers Medium and Long Waves and Trawler band. The ideal radio for home. car ol can be fitted with carrying strap for outdoor use. Completely portable-has built in aeriai for wonderful reception. Special circuit incorporating 2 R.f. stages. push pull output. 3in. speaker (will drive larger speaker), Size $71 \times 5 \% x$ 11n. (Uses PP\& battery available anywhere.)
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Fully tunable over medium and Long Waves and Trawler Band. Incorporates Ferrite rod aerial. tuning condenser, volume control, new type fne tone super iynamaic speaker etc. Attractive case. sizeaker crille. (Uses 1289 battery available anywhere).

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## POCKET FIVE

07 stages- 5 transistors and 2 diodes.
Covers Medium and Long Waves and Trawler Band, a feature usually found in only the most expensive radios. On test Home. Light. Luxembourg and many Continental stations were recelved loud and clear. Designed round supersens1tive Ferrite Rod Aerial and fine tone 2ain. moling coll speaker, bullt intoattractlve black case with red speaker wrilie. Size $5 i x$ is x $31 n$. (Uses PP 4 battery avaliable any. wheres.
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4216
P. \& P. 3\%-
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Listen to stations haif a world away halt a world away with this 5 waveband
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OH： | 8／－ | ${ }_{4} \mathrm{BBP}^{7}$ | 25／． | 6 V 81 | $3 / 9$ | 20 P 514 |  | EACP1 | 3／6 | 81,37 $17 / 8$ <br> 1.48  <br> 126  | KTWH？ $8 / 6$ |  | 8／3 | Cti |  |  |
| O240 | 4／8 | ४8и\％ | $7 / 6$ | か6at | 61. | ${ }^{23,469} 776$ | （1）17／8 | EAR「せ | $7 / 6$ | 61， 318 | hTW63 619 |  | ${ }^{6} 4 / 11$ | Ctí | ／8 | Q13 5 5／8 |
| 1．：3 | $2 / 8$ | ${ }^{6} \mathrm{BW} 7$ | 51. | uxt | 41. | 25L6CT 6／8 | ACIPEN | R B34 | $1 /$ | E141 7／6 | KTV41 61－ | R18 | $34 / 118$ | U50 | $4 / 8$ |  |
| 1A5 | 31． | $\mathrm{HBXB}^{\text {d }}$ | $4 /$ | 6x | 4／6 | 25U4GT18／2 |  | EB41 | $4 / 9$ | ELG3 $\quad 89$ | Lesion 61－ | H18 | 14／． | U70 | 4／9 | （1ET102 8／8 |
| 1a7G | 819 | $6{ }^{6} 4$ | $2 / 3$ | 6 Y 6 | \％1－ | $23 y 380$ | AC／SG 22／8 | E891 | 2／8 | ELS 818 |  | R19 | 14. | U78 | 4／－ | GET103 8／－ |
| 1 Cl | $4 / 9$ | 605 | 4／－ | 786 | $9 /$ | 2515 g 8 8， | AC／EG／VM | EBEC33 | 20／8 | ELL84 516 | LV\＄319 1018 | R5． | $\%$ | U84 | $47 / 7$ | GET10410－ |
| 102 | $8 / 9$ | bcb | 31－ | 787 | 718 | ${ }_{2}^{2524} 1$ | ${ }^{\text {core }} 13814$ | EBC33 | 878 | ELas ${ }_{\text {Elis }}$ | ${ }_{\text {LP3 }}$ | RS | \％ | U101 | 19／6 | GET10s1z／8 |
| 1C3 | 6／9 | 60 | 3 － | 7 Cs | 101－ | 2325 8／－ | ACTH1 $38 / 4$ | EBC41 | ${ }_{71}{ }_{1}$ | ELas $7 / 9$ | ME41 16／10 |  | 64／． | U107 | 17／6 | GETIH12／－ |
| 1 Cs | 818 | ${ }^{6} \mathrm{C}$ | $11 /$ | 708 | 718 | ${ }_{2780}^{28260 T} 83 / 8$ | AC／TP ${ }_{\text {AC／VPI }} 120$ | EBC91 | 8／8 | ELSO | ME91 $12 / 8$ | RK34 | $7 / 0$ | U191 | 11／． | GET113 8／－ |
| ${ }_{106}$ | $10 / 6$ | ${ }_{6}^{6 C 10}$ | 719 | 7D3 | 21. | $\begin{array}{cc}2784 \\ 2857 & 23 / 8 \\ 7 /\end{array}$ |  | E．BF90 | ${ }_{7} / 8$ | $\begin{array}{ll}\text { ELg } & \text { E／93 }\end{array}$ | MH4 3／8 | 8180 | 22／8 | U231 | $9 / 8$ | GET114 $6 / 6$ |
| 105 | \％－ | ${ }^{6012}$ | 8／6 | 7D3 | $161-$ | $\begin{array}{ll}28 D 7 & 7 /- \\ 3001 & 6 /-\end{array}$ | ACYP2 ATP4 $2 / 8$ | EBF83 | 81. | ELaso 27／－ | MHD4 8／6 | SPsb | 19／8 | L281 | $9 / 8$ | GETET2 10／－ |
| 106 | $8 / 9$ | ${ }_{6017}$ | 12／6 | 7D6 | 16／－ | 30 Cl $6 /-$ <br> 30 Cl $10 \%$ | ATP4  <br> $\mathrm{AZ1}$ $8 / 8$ <br> $8 / 8$  <br> 18  | EBFFM | $7 /$. | ELS20 18／2 | MHL4 \％／6 | splis | 12／6 | U28： | 14／8 | GETE73 9／3 |
| 115 | 613 | 6 CD 6 | 21／6 | 7D8 | 15／\％ | ${ }_{30 \mathrm{Cl}}^{30} 10 /-$ |  | EBL21 | ）－ | EL822 19／8 |  | AP41 | $2 /-$ | U301 | 121－ | GETAT4 $9 / 6$ |
| 178 | $2 / 9$ | ${ }^{60} \mathrm{CHO}_{6}$ | 51. | $7 \mathrm{B7}$ | 5 | 30 Fs |  | ECAL | 4／6 | ELLx0 $20 / 5$ |  | 8 Pr 2 | 12／6 | U329 | $9 / 6$ | GEX36 3／－ |
| 153 | $2 / 9$ | ${ }_{60} 6{ }^{\text {C／}}$ | 24／－ | ${ }_{787}{ }_{7}$ | $12 / 8$ |  | $\begin{array}{ll}\text { A241 } \\ \text { B36 } & 6 / 8 \\ 6 /-\end{array}$ | ${ }_{\text {ECS }}$ | 18／6 | EM14 $17 / 9$ | ${ }^{\mathbf{M}} \mathbf{1} \times 1 \mathrm{~B}$ 22／8 | gPbl | 21 | U339 | 11－ | GEX $3810 \%$ |
| ${ }_{17}^{1 F}{ }^{15}$ | 15／9 | 6121 | 1／8 |  | 14／－ | $30 \mathrm{Ll}{ }^{\text {3 }}$ |  | ECSt | $8 /$ | EM34 8／9 | M8P4 12／－ | 8U26 | $27 / 2$ | U403 | 10／2 | GEX48 8／8 |
| 106 | 81. | 6D6 | 31－ | 8D2 | 2／8 | 30 L 16 10／3 | Cl 12／6 | EC70 | 12／6 | EMs5 12／－ | MU12／14 $6 /-$ | sU6L | $6 / 3$ |  |  | GEX64 11／6 |
| 1HSOT | 8／3 | 6D8 | 151． | 8 D 9 | 3／－ | 30P4 12／3 | C10 1816 | EC81 | $27 / 8$ | FM71 22／8 |  | T 1 | － | U8 | 81． |  |
| 114 | $2 / 8$ | 6E5 | 8／－ | 9BW | 13／6 | 30P12 78 | CBL1 $12 \%$ | ECPO | $8 / 8$ | EM80 6／8 | $\begin{array}{ll}\text { N37 } & 23 / 8 \\ \text { N78 } & 2612\end{array}$ | TDD9 | 12／6 | V3t | 18／2 | Mat100 819 |
| llab 16 | 16／10 | ${ }_{671}$ | 9／8 | ${ }^{9 D}{ }^{\text {2 }}$ | 3／－ | ${ }^{30 \mathrm{P} 16}$ 61－ | CCH35 $13 /-$ | ${ }_{\text {EC91 }}$ | $8 / 9$ | EM81  <br> EMY4 813 | ${ }_{-} \mathbf{N 7} 108{ }_{26 / 2}^{26 / 2}$ | TH21C | 14／6 | VM8 |  | MAT120 $2 / 9$ |
| 1LD5 | $4 / 3$ | 6 F 5 | b／8 | $9{ }^{9} 7$ | $12 / 3$ | ${ }_{30 \mathrm{Pl1}}^{30 \mathrm{Pl}}$ 12／8 | CL4 $83 / 10$ | ROC91 | $7 / 8$ | EM85 $9 / 3$ | N118 29／8 | THsoc | $14 / 6$ | $\checkmark \mathrm{P} \cdot$ | $3 / 6$ | 3AT121 $8 / 6$ |
| 1LNS | $4 / 6$ $8 / 6$ | 6F6G <br> 6F6G | ${ }_{7 / 6}^{4 / 8}$ | 10C1 | 9／8 181 － | ${ }_{30 \mathrm{PLL1}}^{30 \mathrm{PL}}$ 10／－ |  | ECC32 | 4／－ | EM97 1512 | N339 15／－ | TH41 | 181． | $\mathrm{VP2}^{2} 8$ | $9 / 6$ | OAS 8／－ |
| 1 Pl | $6 / 3$ | ${ }^{6 F 8}$ | 81. | 10D1 | 71 | 30PL14 19／3 | CV6 $2 / 6$ | ECCO 4 | 21／： | EN31 71／－ | $\mathrm{Pb1}^{2}{ }^{2 / 9}$ | TH233 | $17 / 6$ | VP4 | 14／8 | 10 81－ |
| 1 P 10 | $4 / 9$ | 6F11 | 17\％ | 10D1 | $11 / 8$ | $354580 / 9$ | OV88 $10 / 6$ | ECCC4 0 | ${ }_{7 / 8} / 8$ | $\begin{array}{ll}\text { EN91 } & 5 / 6 \\ \text { EYo1 } & 8 / 3\end{array}$ | PABC80 PC86 $11 / 6$ | ${ }_{T P 2}$ | 6／\％ | VPs | $28 / 8$ | $\begin{array}{ll}\text { OAFO } \\ \text { OA－3 } & 3 /- \\ \text { 3／－}\end{array}$ |
| ${ }_{18} 11$ | ${ }^{6 / 8}$ | ${ }^{6 F 12}$ | 3／－ | 10F1 | 10／－ | 35 LbGT 35 Wa $6 /-$ | $\begin{array}{lll}\text { CV271 } & 18 / 8 \\ \text { CV12 }\end{array}$ | ECC81 | $4 /-$ | EY81 81－ | PCRB $14 / 7$ | TP262 | 17／6 | VP13C | 71. | OAio 8\％－ |
| 1 R | 49 | ${ }_{6 F 14}$ | 25／21 | 10F18 | 101. | 35231814 | CYiC 18／2 | ECC82 | 4／6 | EY83 11／8 | Pc9s 13／－ | TY86 | 11／6 | $V^{1} 23$ | $2 / 8$ | OA81 3／－ |
| 185 | 8／9 | 6 F 15 | 8／8 | 10 LD 3 | 6／8 | $3524 \mathrm{GT} 4 / 8$ | CY91 8／9 | ECcas | 4／8 | EY84 14／－ | Pcy ${ }^{8 / 9}$ | UABC | 18／8 | VP4 | $5 / 6$ | OASS 3／－ |
| $1 \mathrm{~T}^{2} \mathrm{z}$ 3 | 34／11 | 6 F 16 | \％ | 10LD1 | 0／9 | 352507616 | D1 $1 / 3$ | ${ }_{\text {ECC84 }}$ | 6\％ | EY866 6／0 | PCCOS ${ }^{\text {P／8 }}$ |  | 8／－ | ${ }_{\text {VP133 }}$ |  | OAs8 4／－ |
| 174 | $2 / 9$ | 6117 | $12 / 6$ | $10 \mathrm{P13}$ | 818 | 408UA 18／2 | D15 $\begin{array}{cc}13 / 8 \\ \text { D } 41 & 3 / 3\end{array}$ | ECcs | 11－ | $\begin{array}{ll}\text { EY88 } & 9 / 3 \\ \text { EY91 } & 3 /-\end{array}$ | Pccss $11 / 8$ | UBC | 18/2 | VR105 | 8／6 | OA99 ${ }_{\text {O191 }}$ |
| $1{ }^{14}$ | $7 \%$ | ${ }_{6}^{6 F 18}$ | 13／5 | 10P14 | 12／－ | $418 \mathrm{BTH} 2 / 11$ | D41 $8 / 3$ <br> D4 2 $10 / 6$ | EC | 11／－ | Ex9 $3 /-$ <br> cza  <br> $1 / 8$  | PCC8\％ $7 / 9$ | UBCs 1 | $17 / 1$ | VR180 | 5／． | 0496 3／8 |
| 1U3 | ${ }^{6 / 3}$ | ${ }_{6819}^{8519}$ | ${ }_{9 / 8}^{5 / 8}$ | 11D3 | 17／8 | $\begin{aligned} & 42 \\ & 43 \\ & \hline 10 /- \end{aligned}$ | $\begin{array}{ll}\text { D43 } & 10 / 6 \\ \text { D63 } & 5 /-\end{array}$ | ECCP07 | 22／6 | EZ40 819 | PCC189 10／6 | UBF80 | \％ | Vteia | 71 － | OA310 8／6 |
| －2A7 ${ }_{2026}$ | 1018 3 ／－ | ${ }^{6 \mathrm{~F} 23}$ | 9／8 $11 / 8$ | l1D5 | ${ }_{151}^{17 /}$ | ${ }^{43}$ SOA5 $21 / 10$ | ${ }_{\text {D77 }}$ 2／3 | ECF80 | 8／8 | EZ41 8／8 | PCF80 6／－ | UBF＇8 | 713 | VTsil | 3／－ | OA211 13／6 |
| 2 D 130 | $\%$ | 6F22 | 81－ | 11E3 | 17\％ | 508б 7／． | DAC33 8／3 | ECF82 | $7 / 8$ | E780 4／6 | PCF82 $8 / 8$ | UBL2 | 11． |  |  | CClBw 351－ |
| 2D－1 | b／6 | $\mathrm{HFH}_{518}$ | 12／3 | 12A6 | $2 / 3$ | s0cs 7／－ | DAFY1 3／8 | ECFE | 19／5 | 281413 | CF84 14／7 | HCy | 7／6 |  |  | 251－ |
| 2 P | 23／3 | $6 \mathrm{~F}^{2} 2$ | 4／． | 12A8 | 18／8 | 30CD6G4019 | Dar96 $8 / 3$ | ECF8 | 201． | E290 4／－ | ${ }^{\text {PCFP8 }}$ PCLS ${ }^{\text {\％}}$ | UCXPs | ／－ | VU133 |  |  |
| 2 X 2 | 81－ | ${ }_{8}{ }^{8} 3$ | 9／8 | 12 cce | 13／5 | ${ }^{50 L 6 G T} 7 /-$ | DCC90 ${ }_{\text {D／4／9 }}$ |  |  |  | PCLS3 $8 / 8$ | Hors | 10／8 | W21 | 5 | $\begin{array}{ll}\text { O23 } & 181 \\ \mathrm{OC}_{3} & 12 \%\end{array}$ |
| 3 A 4 | 4. | $6{ }_{6} 6$ | $2 / 8$ | 12AD6 | 13／2 | ${ }^{52 \mathrm{KGU}} 14 / 6$ | $\begin{array}{lll}\text { D14 } & 19 / 8 \\ \text { DD41 } & 12 / 8\end{array}$ | ECH33 | 22／8 | $\mathrm{FCl3}^{\text {c }} 17$. | PCLA4 $5 / 3$ | －CH21 | 1 9／－ | $W_{42}$ | 22／8 | OCy6 25\％ |
| ${ }^{3145}$ | $8 / 9$ 51 |  | 1／8 | 12AE6 $12 \mathrm{AB7}$ | 12／3 | $\begin{array}{cc}\text { 63EG } \\ 7 \% & 14 / 6 \\ 8 / 6\end{array}$ | $\begin{array}{lll}\text { DD4 } & 1888 \\ \text { DDT4 } & 8 / 6\end{array}$ | ECH35 | 8／6 | FW4／5008／8 | PCL＊5 8／8 | UCH42 | $27 / 3$ | W61） | $27 / 3$ | OC28 12／6 |
| $3 \mathrm{D6}$ | 4／－ | 6J5a | $4 / 3$ | 12A号 | 91. | 78 5／6 | DLT25 $7 / 6$ | ECH42 | 718 | FW4／8008／8 | PCL＊ $10 /-$ | 1 CH | 91 | ${ }^{4}$ ¢ 3 | $10 / 6$ | OCas 8 87／6 |
| 34 | 5／8 | 8J6 | 3／－ | 12AT8 | 4／0 | is 12／6 | DF93 8／6 | ECH81 | 618 | OT1C 101. | PULs8 12／6 | CCl | 8／8 | ${ }^{4} 78$ | 3／8 | 35 c |
| 945 | $7 / 3$ | 8J79 | 4／9 | 12AT7 | $4 /$. | 77 8／－ | Fion 151－ | ECH83 | 817 | GUb0 41／8 | PEN4DD | UCLI | 9／8 |  | $2 / 8$ | OC36 21／8 |
| 384 | 4／9 | 6.769 | 7\％ | 12．aU8 | 816 | 78 5\％＊ | DF72 301－ | ECH84 | $14 / 7$ | G230 \％／－ |  | UF4 |  |  |  | OC41 |
| 344 | 8／8 | ${ }^{3} 88$ | 12／8 | 13aU7 | 4／6 | $80 \quad 6 / 8$ | DF91 2／8 | ECL80 | 6／－ | 67822 ${ }^{\text {278 }}$ |  | UFP0 | 6／8 | W107 | 2015 |  |
| $4 \mathrm{D1}$ | 4／－ | 6K6 | $8 /$. | 12ave | 8／8 | 83 15／＊ | ${ }^{\text {DF96 }}$ 8／3 | ECLs\％ | 719 | $023317 / 6$ | － | UFSS |  | W729 | $17 / 6$ | $\begin{array}{ll}\text { OC43 } & 12 / 8 \\ \text { OC4 } & 8 / 8\end{array}$ |
| 8R4G | 91. | 6K7G | 1／6 | $12 \mathrm{~A} \times 7$ | 4／9 | 89V 19／5 | UF97 $7 / 8$ | ECL83 | 10／． | $\begin{array}{ll}0234 & 11 / 8 \\ \text { G237 } & 14 / 6\end{array}$ | PEN45 84／6 | UFs＊ |  | \＄14 | 186 | OC44 OC4 PM 9／3 |
| 5 T 4 | $8{ }^{1 /}$ | 6K70T | $4 / 8$ | libab | 8／8 | 80al 55／8 | DH\％15／8 | EFb | 90／－ |  | PENTODD | UF9 | 6／3 | X18 | 619 | OC4PM OC45 O／3 |
| $\mathrm{SCH}_{4}$ | $4 / 3$ | 6K89 | 4 \％ | 12BE6 | $6 /$ | 85 AL 9／9 | $\begin{array}{ll}\text { DG63 } & 4 / 8 \\ \text { DH76 } & 3 / 9\end{array}$ | EFFi | 20／6 | ${ }_{\text {H80 }}$ H83／－ | 46D $16 \%$ | UL41 | \％－ | $\times 24$ | 16／8 | OC45 OU45 9／－ |
| 5F4G | 7／6 | 6 KgGT | $7 / 9$ | 12 BH 7 | 718 | 9019 90476 | $\begin{array}{ll}\text { DH76 } & 8 / 8 \\ \text { DH77 } & 4 / 8\end{array}$ | EF2\％ | 8818 |  | PEN46 4／6 | UL4 | 23／3 | X 41 | 23／3 | $0_{0} 0$ |
| ${ }^{51830}$ | $4 / 8$ | 6 K 26 | 13／6 | 12 EL |  |  | $\begin{array}{ll}\text { DH77 } & 4 / 6 \\ \text { DH81 } & 8\end{array}$ | EF36 | 8／3 | 101－ | PENis311／8 | UL4 | 91－ | ${ }^{1} 61$ | 10／－ | O636 |
| 3Y4 | $9 / 6$ | 61.1 | $9 / 8$ | 12.2080 | 1／18 | $\begin{array}{ll}9000 & 37 / 8 \\ 90 . & 42 / .\end{array}$ | DH10127／11 | E．F97A | 61. | HL2 7／6 | PES453D： | ULA4 | 6／3 | X69 | 7／－ | Oc70 8／8 |
| 523 | $19 / 5$ | ${ }_{6}^{61 . \%}$ | 818 | $12 J 50$ | 2／6 | 90 cl 15\％ | DH107 1818 | EFsy | $8 / 9$ | HL13C 4／－ | 17／8 | UM4 | $15 / 2$ | X $\mathrm{x}_{5}$ | $4 / 6$ | Oc71 8／8 |
| ${ }^{52 / 30 \mathrm{La}}$ | 81／8 | ${ }_{6 L 70 T}^{6 L 6 M}$ | 4／8 | 12 K 5 | ${ }^{17 / 18}$ | ${ }_{150 \mathrm{~B} 2}^{00} 16 / \mathrm{l}$ | DK32 8／9 | EF40 | 101． | HL23 14／11 | PENA4 $7 / 8$ | CM34 | 18／10 | ${ }^{\times} 65$ | 516 | OCJ 8／－ |
| CAi | $91-$ | 6L17 | $12 / 6$ | 12kic | T 318 | $150 \mathrm{Cl}^{2} \mathrm{~S}$ 5／－ | 1以 40 18／8 | EF＋1 | 7. | HLa3LD 5／－ | PEN | Cstay | 9／3 | X86 | \％／6 | Ocis 18／－ |
| Ba8G | 7. | ${ }^{6} 1.18$ | \％／8 | 12 Kmot | 91－ | 161 18／－ | DK91 4／9 | EFs？ | 8／3 | HL＋1 3／9 | 23／11 | 1110 | 71 | －774 | 11／－ | OC7 +81 － |
| 6487 | $4 /-$ | 6L19 | $9 / 9$ | 12476T | $3 / 9$ | 186BT 34／11 | Dh93 $0 / 8$ | EFbo |  | HL420 | PEN／DIM | CUS | 71. | X 78 | 2818 | OC73 8／－ |
| BAC7 | 3／－ | 6LD3 | 7. | 128 A ？ | 7 | 2153G 8／4 | DK9 8\％8／8 | Brit． | $1 / 6$ | $8 / 6$ | $1 \pm 0{ }^{1718}$ | UU6 | 11／9 | \＄79 | $40 / 8$ | O以76 8／6 |
| 6atis | 2／9 | $6 \mathrm{LH13}$ | 71. | 12307 | 41． | $220 \mathrm{~B} \quad 10 / 6$ | DL3s 7／3 | Ame | 2／6 |  | ？133 1811 | UU8 | $12 / 8$ | － | ${ }_{23 / 8}$ | C7\％12i－ |
| 6 6a7 | $8 /$ | 6L．L20 | 12／6 | 12807 | 3／． | 301 201－ | D1，35 $6 / 8$ | $\mathrm{EFP}^{\text {P }}$ | $3 /-$ | HN309 26／8 |  | UU9 | $12 / 8$ $5 / 6$ | X109 | $29 / 1$ | OC78 8 8／－ |
| cajs | $8 / 8$ | $6 \times 70$ | 5 － | 12887 | 3／－ | $30210 / 8$ | DLAS $61 /-$ | EF73 |  |  | PLSI $23 / 3$ | CU19 | 4／8 | $\times 118$ | 9／6 |  |
| 6AK5 | 8／． | ${ }^{6 P 1}$ | 18／日 | 12817 | 51－ | ${ }_{303}^{303} 10 / 8$ | $\begin{array}{ll}\text { DLAA } & 17 / 6 \\ \text { DLAB } & 151\end{array}$ | ${ }_{\text {EFr83 }}$ |  | HVR2A \％\％ | ${ }^{\text {PLAI }}$ S／B | UYIN | $10 / 9$ | $\times 119$ | $2 / 3$ | 0c82 10／－ |
| 6AK ${ }^{\text {c }}$ | 12／6 | ${ }_{6 P 25}$ | $8 / 5$ | 12\％K7 |  |  | $\begin{array}{ll}\text { DLA8 } & 15 \% \\ \text { DL．72 } & 15 / \%\end{array}$ | EF\％ | 8. | ${ }_{1 \times 3}{ }^{\text {E／8 }}$ | PLA5 $8 / 6$ | UY21 | 8／9 | $\times 142$ | 7／3 | OC83 8\％ |
| 6aKs | 8／3 | $6 P 28$ 6 P 28 | $19 / 6$ $11 / 8$ | 12807 12467 | 8\％－ | $\begin{array}{ll}303 & 18 / 2 \\ \$ 0+1 \\ 3 & 13 \%\end{array}$ | $\begin{array}{ll}\text { D1．72 } & 151- \\ \text { D175 } & 301-\end{array}$ | EF86 | 8／6 | 1w ${ }^{\text {W }}$／R50 $5 / 8$ | $\begin{array}{ll}\text { PL／44 } & 8 / 6\end{array}$ | UY41 | 8／－ | Y63 | 81－ |  |
| 6ams | $2 / 6$ | 6Qto | $4 / 6$ | 12 Cba | $\%$ | 886A 12／6 | D L9：2 4／9 | EFg9 | 419 | 1154／5int $8 /$－ | P1，820 18／4 | UY35 | 4／6 | \％ | 51. | OC139 13／6 |
| 6asic | $3 /$－ | 697CT | $7 / 9$ | 12 Y 4 | 9／－ | 4083 15／－ | D1．94 $51 / 8$ | Er91 | $3 /-$ | KBC32 20／5 | PM84 $9 / 6$ | ${ }^{\text {C10 }} 10$ | $8 /$ | ${ }_{786}$ | 19 | OCl40 19／． |
| 6AU： | $8 \%$ | 8R－0 | BJ－ | 148n | $20 / 9$ | 4887 71／－ | D1．45 6／－ | EFP3 | 2／6 | KF3s 1218 | 10／－ | ， | $7 / 6$ | －7 | $81-$ | $0 \mathrm{Cl7}{ }^{818}$ |
| 8 AH6 | 201－ | 6Rigt | 11. | 14137 | 29／8 | 5763 7／6 | DLat 6／3 | EF9， | 5／－ | K138 $12 / 6$ | ${ }^{10} 6$ | U | $10 / 8$ | ${ }_{7}^{2.319}$ | $81-$ | $\begin{array}{ll}0 \mathrm{Cl} 171 & 10 / 6 \\ 06200 \\ 10 / 6\end{array}$ |
| ${ }^{\text {bato }}$ | $4 / 6$ | 68A7 | $6 / 9$ | 1487 | 17／8 | $\begin{array}{ll}7193 & 1 / 8 \\ 7478 \\ \text { 3／－}\end{array}$ | Dimio $\begin{gathered}\text { 10／8 } \\ \text { Dw } \\ \text { 5\％}\end{gathered}$ | EFY8 | $11 / 8$ | ${ }^{\mathbf{K} T 2} \mathrm{~V}$ \％／6 | PY゙y2 $9 /$ | C＇1812 | 816 | 7719 | 4／－ | $\begin{array}{ll}\text { Oc200 } & 10 / 6 \\ \text { OC201 } \\ 36 /-\end{array}$ |
| ${ }^{\text {OAUG }}$ | $5 / 8$ | 6847 6867 | $4 / 9$ | ${ }_{18}^{15}$ | ＋6／－ | A1834 20\％ | ${ }_{\text {DM71 }}^{\text {Dw70 }}$ | EF163 | 9／0 | KT8 15／ | PY33 10／－ | U19 | $48 / 6$ | 2729 | 8／6 | Oc203 14／－ |
| ${ }^{6} \mathrm{SAV} \mathrm{S}^{\circ}$ | $5 / 6$ | 6867 6847 | 4／9， $3 /$. | 18 | $12 / 8$ $10 / 8$ |  | DW4／3508／6 | Erist | 819 | $\begin{array}{lll}\text { KT32 } & 8 / 8\end{array}$ | P180 5／9 | $\mathrm{Ul}_{1}$ | 151 |  | $10 / 6$ | OCP71 17／6 |
| 6B8G | $2 / 6$ | BES： | $6 \%$ | 19 AO 5 | 719 | AC2HL $10 / 8$ | DW4／5008／6 | EF804 | $20 / 5$ | KT33C 4／－ | PY81 819 | ${ }_{\text {U } 28}$ | ${ }^{6 / 8}$ | 27 | 36／－ | ORP12 18／6 |
| 6BAB | 5\％ | 68 K 7 | 4／8 | 19 BGBG | 020／5 | AC2PEN | DY86 71－ | EE90 | 15／2 | KT96 32／4 | PYgiz ${ }_{\text {PY83 }}$ | U94 | $15 /-$ |  |  | 8X841 10／－ |
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HI-FI EQUIPMENT CABINETS
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Model 882 witb BALANCE CONTROL $110 / 250 \mathrm{v}$. A.C. imput. 5 watt undistorted output ( 10 watte nombnal). Size $12 \times 9 \times 2 \mathrm{in}$. Weight 9 lb . Complete with spec. and instructions. STILL ONLY 25.19.6 Carr. 7/.

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E．M．I．4－speed
Player and P．U． further huge oftor these $67 / 6^{P}$ ，


Henfy Blin．metal turn－ farmese 200 fintiter per． motar mith tap at $4 a \mathrm{~V}$ for amplifier valve flament if regitired．Turnover LP／78 head．
RECORD PLAYER AMPLIFIER 2 Falve（EZSO，ECLS2）． output，ready built．tested and complete with ratves and outpint transformer． gize in．w．x 2bla．d．$x$ 5tn．h．55／－．P．\＆P．3／～． Suitable speakers：Rin．15／－ $\underset{P}{P}$ \＆P，1／6． $10 \geq 61 \mathrm{~m}$, ， $85 /-$
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3 l．F．tranaformers，one oncillator coil，one drives transormer and wonnd Fertite aerial（med．，long a transistor printed circuit．borid to match， $8 / 6$ ， post 9d，Circuit diagram $1 / 6$ extra，

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AMPLIFIER
A top－quallty revord player amplifier．This
 and volume controla．Complete with output trans． loriner instchen sor 3 ohms speaker PRICE 6916．P．\＆P． 316
DITTO．Molnteil on board with output tranb－ iormer and fin．apesker． 896, P．\＆P． 416.
Complete at $89 / 6, ~$

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GET 15（Matched Pair）15／

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21／n．12／6；5in．12／6；A立㶽．15／－； 8 in． $21 /-101 \mathrm{n}$ ． $25 /-$ ；12tn． $27 / 8$. 8in．I Sin By famous 10／6 maker ．．．．．．10／6 E．M．I． 13 \} ln. r 8 tin．high flux $\quad 82 / 6$ oin．midule register sueaker

$$
\text { mao ls ohm } 12 \text { lach., } 30 /
$$

P．\＆P．up to 6ln．1／6：over fin．
2／6 per spualer．
AMPLIFIER ON

## PRINTED CIRCUIT

## BOARD

O．P．trans．，uge U8，ULSA sp of motor use with 80 volt tap of motor， $39 / 6$ ．
P．P．2／6 on sbove，Dropper rea．for fllamanta if required．

## B．S．R．AUTO UNITS

 160 ．Suitable for use withabove．（ Slightly soiled．） 24.4 .0 ．
 with 8 ohm apeaker．se．g．6．Carr．B／
Superior CABINET gimllar to ahove to take 8 x gin pesker with motor hoard，will accommodate BAR UAl4 or UAl6．\＆3，9．8．Carr．B／6． spenkur $15 /$－extra．P．\＆$P \cdot 1 / 6$ extra．

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ELECIROSTATIC H．F．IWEETERS．Type L．S．LI．is size 3 x 3 in．， $2 / 6$ MIDGET 2／GANG CON
（ case with bulit－in trimmern．Size i i i itin．Not used but removed from P心 Boarde．Two for $8 /-$ ，plue $1 /-P$ \＆$P$ ． ACOS CRYSTAL WIKES．High imp．For deak or hand ure．Bigh TBL CRYSTAL STICK MIKE
TR Men TRANEISTOR DRIVER and O／P TRANSFORMERS．（Tapped 3 ohm hid 15 ohnas output）．Wha + Endable Tranabior gitning appros， 1 suth outnut 30／－
 ACOS GPBG／1 T／O MONO CRYBTAL CARTRIDGE．Complete ailh apphire wi，lia and mountlay bracket．Simited number nuc 81 12／6． F ，\＆ P

TAPE DECKS
COLLARO STUDIO DECK E10．10．0 plua 5／6 carr．and ins B．S．R．MONARDECK

 （Tapeesestra on bot $h$ ）．

SPEAKER \＆ CABINET FABRICS Oatmpal fabric for men． ker ir cmbinet or Red rexine fir cabinet sain． Irnath．（Minimam order） P．\＆$P$ ．$J / h$ ．

# Stow <br> <br> MULLARD 3-VALVE <br> <br> MULLARD 3-VALVE PRE-AMPLIFIER TONE PRE-AMPLIFIER TONE CONTROL UNIT CONTROL UNIT <br> <br> esigned mainls for Mullard 

 <br> <br> esigned mainls for Mullard} CUULARAD DESTGNS COMPLETE KITS OF PARTS

Range of Amplifiers. also suitable forany Amplifiers requiring input up Channels. including for Tape and Magnetic Pickups. Separate Bass and Treble controls. High pass filter 20 to $160 \mathrm{c} / \mathrm{s}$. low pass filter $5-9 \mathrm{Kc} / \mathrm{s}$. Totally enclosed in case size $11_{1^{*}} \times \mathrm{x} 41^{\circ} \mathrm{x} 4^{*}$ KIT OF PARTS \&10.0.0 ASSEMBLED \& TESTED \&13.13.0 (Carr. \& Ins. $5 /$-).

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For use with MULLARD 2 or 3 valve preamplifiers with which an unuistorted power output of up to 10 watts is obtained. SPECIFIED COMPONENTS AND MULJ,ARD VALVES including PARMEKO MAINS TRANSFOIRMER and choice of PARMEKO or PARTRIDGE Output Transformer.
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The popular complete " 5 -10" incorporaling Passive Control Unit providing up to 10 watts high quality reproduction with and new MULLARD VALVES. Includes PARMEKO MAINS TRANSFORMERS and choice of PARMEKO or PARTRIDGE Output Transformers. Surplus Power available for Tuner.
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A HIGH QUALITY AMPLIFIER DEVELOPED FROM THE VERY POPULAR 3-WATT MULLARD "3-3" DESIGN. KIT OF PARTS
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Switched inputs for 78 and L.P. records plus a Radio position. Extra power to drive a Radio Tuning Unit is also available. (Carr. \& Ins. 6/6). Please state L.S. impedance.


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Perfectly suited for Portable Installations for Which purpose we offer PORTABLE CASE SPEAKER ( $£ 1.0 .0$ ). All for
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Alternatively with ASSEMBLED AMPLIFIER.
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Employing two EF86 valves and designed to operate with the Mullard AMPLIFIERS but also perlectly suitable for other makes with input up to $250 \mathrm{~m} / \mathrm{V}$.
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KIT OF 86.6 .0 ASSEMBLED 89.10 .0 (Carr. \&
PARTS PRICE REDUCTIONS
(a) THE KIT OF PARTS to build both the ${ }^{15} 5-10$ Amplitier and t
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£3.10.0 speed autochanger with crystal pick-up
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 Single Record Player fitted with the latest T.P.A. 12 pick-up arm and G.C.8. crystal Cartridge

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Glarkard monel S.R.P. 10 single record player fitted with high output crystal pick-up can be operated both manually and automatically. Sultable for Mono or Stereo operation
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PARTS
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A self contained Unit consisting of Garrard Deck and matched Preamplifier on one chassis. Provides full tape through pyck Up Sockets of Standard through Pick Up Sockets of Standard
PRICE tncludes complete Tape Marazine. £18.18.0

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STEREO 12 Mk 2
840／5／0
8 watts push－pull output from each channel． 16 Watts total， medium and long banda；A bi－fl system on one
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Junlor version of stereo 12： 5 watts per channel， 10 pick－upe and future stereo radio．
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 R 5 | 6／－らN8 | $5 /$. | ERC41 |  | PCIS4 | 10\％ |
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| 174 | 3／－6NTM |  | FRFF80 |  | P1．83 | 8\％ |
| 2x＇s | 2f－1n47g | $61-1$ | ECH42 |  | PYi33 | 15／－ |
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| 6BA6 | 71.12 K 8 | 14／－ | EZ40 |  | UBF89 | $9 \%$ |
| 6BE6 | 5／－1207 |  | EZ80 |  | U＇H81 | O\％ |
| 6BW6 | 7－25Y5G |  | EZ81 | 71－ | UCLAR2 | 10\％ |
| 6 C 4 | \＄J－351．6 |  | HABC80 | 10／－ | UCL83 | 12\％ |
| 6D6 ${ }^{*}$ | 5i－38Z4 | 5／－ | HVR2a | 5\％ | UF89 | Of－ |
| 646 ？ | 4／－954 | 2／－ | KT330 | 8／－ | UA1 | $9 \%$ |
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| 6.56 ）＊ | 5／－DK98 | 8／－ | PCC84 |  | UU9 | $7 \%$ |
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$4 / 450 \mathrm{~V} 2 / 8250 / 25 \mathrm{~V}$
$5 / 61^{8 / 600 V}$
$3 / 450 \mathrm{~V}$ 2／3 $500 / 12 \mathrm{~V} \quad$ 3／－ $16 / 450 \mathrm{~V}$
$16 / 480 \mathrm{~V} 3 /-1,000 / 12 \mathrm{~V} \quad 8 /-50 / 450 \mathrm{~V}$

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| $32+32 / 450 \mathrm{~V}$ |  |  |


$25 / 25 \mathrm{~V} \quad 1 / 9.8+8 / 450 \mathrm{~V} \quad 3 / 6 \left\lvert\, \begin{array}{ll}32+32 / 450 \mathrm{~V} & 6 / \mathrm{F} \\ 32+32+32 / 350 \mathrm{v}^{7} / \mathrm{l}\end{array}\right.$ | $25 / 50 \mathrm{~V}$ | $2 /-$ | $8+16 / 450 \mathrm{~V}$ | $3 / 9$ | $50+32+32 / 35$ |
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TELESCOPIC CHROME AERIALS．13in．extending to $43 \mathrm{in} ., 8 / 6 \mathrm{ea}$ Cost Adaptor Plug． $1 / 6$ extra， TRIPLEXERS Bands I，II．III
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BALANCED TWIN FEEDER yd．6d． 80 or 300 ohma． DITTO SCREENED per yd． $1 / 6.80$ ohme only． Wirewound Ert．Spesker Control， $10 \Omega 3 /-25 \Omega$ 6／6．
W1RE－WOUND POTS． 3 WATT．Pre－set Min． TV Trpes．All values to 10 ohms to $25 \mathrm{~K} ., 3 /-$ ea． $30 \mathrm{~K} . \quad 30 \mathrm{~K}, \mathrm{f} 4 \%$（Carbon 30 K ．to $2 \mathrm{meg} ., 3 /-)_{0}$ Wire－WOUND 4 WATM8 Pots．Long
Vahe， 50 ohme to $50 \mathrm{~K} ., 8 / 6 ; 100 \mathrm{~K}, 7 / 6$ ．
Vblue， 50 ohme to 50 K. ． $6 / 6 ; 100 \mathrm{~K}, 7 / 6$.
PHILIPS TRIMMERE． $0-10 \mathrm{pF} ., 3-30 \mathrm{pF} ., 1 /-$
TRIMMERS，Cersmic： $30,50,70 \mathrm{pH}^{3}-9 \mathrm{~d} . ; 100 \mathrm{pF}$ ， $150 \mathrm{pF} .1 / 3 ; 250 \mathrm{pF}, 1 / 6 ; 500 \mathrm{pF}^{2} 750 \mathrm{pF}, 1 / 9$ ． TV etc．TRIMMER， 1000 pF ．with knobs． $2 /-$
RESISTORS．Preterred values， 10 ohms to 10 meg． Hikh Stability．${ }^{\text {I }}$ ，wi， $1 \%, 2 /=$ Preferred values $10 \Omega$ to $10 \mathrm{meg}^{2}$ Ditto $5 \% 10 \Omega$ to 22 meg．． 9 d ． $\left.\begin{array}{ll}5 \text { watt } \\ 10 & \text { watt }\end{array}\right\}$ WIRE－WOUND RESISTORS 15 Watt 10 ohms－ $10,000 \mathrm{ohmg}$
12.5 E to 47 K 10 w ．
$\left\{\begin{array}{l}1 / 3 \\ 2 /- \\ 2 /=\end{array}\right.$
3／0
Volume Controls 80 ohm 80 ABL

Lanear of Log Tracks Long apindles．Midget 5 K ohms to 2 Meg ． L．8．3／－：D．P．，4／6： Ekereo L／B 10／6；D．P．14／6

| Semi－air spacen |
| :---: |
| Stranded core $6 d . y d$. |
| $40 \mathrm{yds}, ~$ |
| $1 / 6$ | $40 \mathrm{yds} .17 / 6$

$60 \mathrm{yds} .25 /$
Fringe Quslit
Air spaced $1 /=y d$.

MAINS TRANSFORMERS 200／250 V．A．C． STANDARD， $250 \cdot 0-250,80 \mathrm{~mA}, 6.3$ tapped 4 v． 4 a．Rectifier， 6.3 v ． 1 a． 5 v． 2 a．or 4 v． 2 a． $22 / 6$ ，ditto， $350-0-3 \overline{0} 0$ MIDGET， 220 v． $45 \mathrm{~mA}, 6.3$ v． 2 a SMALL， $2200-0-220,50 \mathrm{~mA}, 6.3 \mathrm{v}$ ． 2 a ． STD． $250-0-250,65 \mathrm{~mA}, 6.3$ v． 3.5 a ． HEATER TRANS． $6.3 \mathrm{v} .1 \frac{\mathrm{amp}}{\mathrm{m}}$ ． Ditto，tapped 1．4，2，3，4，5， 6.3 v．
 $3,4,5,6,8,9,10,12,15,18,24.30 \mathrm{v}$ ． AUTO TRANSFORMER， 150 w ．
$0,115,200,230,250 \mathrm{v} ., 500 \mathrm{w}$. $10 / 6$
$15 / 8$
$17 / 8$ $17 / 8$
$17 / 8$ $7 / 6$
$8 / 8$ $8 / 8$
$10 / 6$ $\operatorname{amp}_{22 / 8}$ ．．82／6 PARMEKO MAINS TRANSFORMER．Made for ppecial contract．the ratings can eafely be doubled．Cuaranteed 2 years．Primary $0-110$－ $210-230-250$ v．H．T， $300-0-300$ v． 50 mA ．L．T． 6.3 v． 1.8 am k ．Size $4 \times 3 \frac{1}{2} \times 3 i \mathrm{n}$ ．

INTERVALVE TRANSFORMERS， $3: 1$ or $5: 1,9 /-$ O．P．TRANSFORMERS．Heavy Dity 50 ma $4 / 6$ dultiratio，$/ 8$. Multiratio hea d $10 \mathrm{w} ., 15 / 6$ ．Miniature，3S4，ete．， $5 / 9$.
10 w．O．P．matching trans． $3.7,15 \Omega, 12 / 6$. L．F．CHOKES $1.3 / 10 \mathrm{H} ., 10165 \mathrm{~mA}, 5 /-; 10 \mathrm{H}, \mathrm{g}, 85 \mathrm{~mA}$ ， TINNED COPPER wIPE
 32－40．4／6：D．C．C．28，34，3f srg，：ooz．3／6．

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 Mullard Anplifier Manual Muliar Valve Guide，Books 1． $2 . \ddot{3}^{8 / 6}$ $\begin{array}{ccccc}\text { Radio Vaive Guide，Books 1، } \\ 4 \text { or } 5 & . . & . . & . . & . . \\ 5 /-e a c h\end{array}$ Transistor Superhet ikeceivers ．． $7 / 6$ Practical Radio Inside Out Practical Radio Inside Dut Transistor Controlled Models Princinles of Colour TV International lRadio Stations＂16／－

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8 p．4－way 2 wafer long spindle ．．6／6
p． 2 －way，or 2 p． 5 －way long spindie
p． 2 －way or 4 p．3－way long spindle $\quad \cdots \quad 3 / 6$ p．4－way or 1 p．12－way long apindle $\quad 3 / 6$ Wavechange＂MAKITS＂．Wafers avail－ 4 pl， 1 p． 12 way， 2 p． 6 way， 3 p． 4 way． ${ }^{4}$ p． 3 way， 6 p． 2 way， 1 wafer switch， $8 / 6$ ． 2 waier switch，12／6； 3 wafer switch， $16 /-$ additional wafers up to $12, ~ 3 / 6$ each extra：
Tomate switehes，
s．p．
$2 /-9$
d．p．．
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Vol. XXXIX No. 680 OCTOBER, 1963


## BE PREPARED!

AT this time of year, as autumn approashes and the even:ings draw in, most readers will be spending less time in outdoor pursuits and more on indoor activities. And in particular, thoughts of the impending constructional "season" will occupy a good many minds.

For whereas the majority of enthusiasts (quite rightly) enjoy as much of the outdoor life as our capricious climate allows during the summer months, the winter sees a peak in constructional activity that restores the average.

After a reduction of building projects for a month or two, most amateurs look forward keenly to this resurgence of activity which begins to gain impetus as the autumn leaves fall. Many readers, of course, maintain a certain level of practical aclivity all through the vear, but nevertheless there is something about the autumn and winter that presents an irresistible lure even to those who temporarily forsake practical work in the summer.

Perhaps it is the cosy atmosphere of the workshop, den or "shack", with the smell of flux on the soldering iron and the low hum of power packs. Perhaps it is because, shut off from the obtrusive world with the curtains drawn, one can get to work without external distractions and concenirate fully on the work in hand. Or maybe it is because the true enthusiast, though seemingly inert during the summer, has been quietly planning new ideas for equipment and now feels the time is ripe to put them to the test.

Whatever it is, one thing is certain: during the next few months much solder is going to flow, many holes are going to be drilled and many thousands of components are going to be wired into hundreds of different circuits. With, we hope. a large percentage of successes!

But despite the fact that this general pattern follows regularly, we wonder how many are really prepared for the approaching high level of activity. The facilities available to the radio amateur vary enormously from. literally, the kitchen table to the separate permanent workshop-with a wide assortment in between.

Yet whatever the enthusiast's personal circumstances. he would do well to ensure that he is fully prepared to embark on an extended period of constructional work. For instance, is the wiring in the workshop adequate and is it really safe? Have those important extra points been installed? Are all tools in good shape? Are there any gaps in the tool kit? is the spares box in reasonable order? Are the boxes of nuts, bolts, small components, etc., sorted and conveniently housed? In other words-are you ready for working at maximum efficiency and ease?

If somewhere along the line you get a negative answer, now is the time to put things right. Before the season really gets under way-when you will never find the time!

Our next issue dated November will be published on October 4th

$\mathrm{A}^{\mathrm{N}}$ order for a number of electronic aids for blind persons, which were described in these pages a few months ago and which rely on the difference of frequency between an ultrasonic beam of energy and its reflection from nearby objects to enable the blind user to "hear" obstacles ahead, has been received by the manufacturers from the Bureau of Rehabilitation Services of the Department of Education for the State of Kentucky, U.S.A.

Ultra Electronics Limited, who produced the first of the aids for the preliminary evaluation tests, have also received an order for 50 units from St. Dunstan's.

## H.F. Equipment for Ministry of Aviation

'THE Ministry of Aviation has recently ordered a number of Marconi type D11 and D13 highfrequency mobile radio stations. The contract is valued at approximately $£ 1,500,000$.

The type D11 transmitter/ receiver equipment is a mobile field communication unit providing the latest h.f. facilities, including s.s.b./is.s.b. telephony and f.s.k. telegraphy, with incorporated teleprinter. The output power of the transmitter is rated at $300-350 \mathrm{~W}$ over the frequency range 2 - $21.999 \mathrm{Mc} / \mathrm{s}$. The receiver is a high stability independent side-band and telegraph equipment, crystal controlled and operating on the double superheterodyne principle.
The type D13 equipment operates over the same frequency range but with a transmitter power output rating of 1 kW . It also differs from the D11 in that it employs two receivers (for dual diversity operation) instead of one.

## REGULATORS FOR BBC

 ' ${ }^{\text {PHE BBC's v.h.f. transmitting }}$ and radio link stations as well as television stations, require very stable voltage supplies, and recently at the Wimbledon works of Foster Transformers (a member of the Metal Industries Group) six new voltage regulators for various BBC stations have been undergoing extensive tests prior to delivery.Each of the regulators has an output of 17 kVA which is necessarily maintained stable for the transmitter and associated equipment it will be serving when installed.


One of the voltage regulators made by Foster Transformers for the BBC.

## IMPROVED SOUND FOR ST. PAUL'S

AFTER extensive tests made by engineers from Standard Telephones and Cables Limited, the sound system of St. Paul's Cathedral, London, has undergone considerable alteration, designed to improve audibility to every part of the congregation.

The Cathedral's architecture and size have posed special problems for the sound engineers involved on the operation, and in
particular there is a problem of the time lapse of speech from the altar and pulpit areas, and along the length of the Nave. (It was found that sound took threetenths of a second to travel from the altar to the far end of the Nave.)

As it was decided to incorporate the Cathedral's existing system of loudspeakers in the new improvements, it was obvious


An STC engineer positions an 8 ft . column loudspeaker in St. Poul's Cathedral during recent accoustic tests.
that a delay mechanism would be necessary to co-ordinate sound arriving direct from the source and the sound delivered over the loudspeakers; which, in the normal course of events, would arrive almost inmediately. Failure to compensate for this phenomenon would result in garbled speech reaching the distant parts of the congregation.

The method used by STC to give the necessary delay employs a magnetically loaded neoprene belt, stretched around a metal drum. In operation the drum is revolved at 80 revolutions per minute. Several magnetic heads are distributed at specific points around the drum and held in contact with it as it revolves. One of two of these heads. records speech on to the helt from either the altar or pulpit 'microphone. As the drum revolves, the recorded speech is carried to replay heads successively spaced further around the drum from the recording head; and from these the speech is carried to loudspeakers situated at various distances from the microphone. The time taken for the signal to travel round the drum from record to replay heads provides the necessary delay.

Finally as the recorded speech moves past the last replay head, an erase head removes the signal hefore once again the recording heads are reached at the end of one complete revolution of the drum.

## SERVICING EXAMINATION

IT was announced recently that out of the 853 candidates who entered the final written examination for the Radio and Television Servicing Certificate of the City and Guilds of London Institute, 456 passed and thus qualified for the practical test to be held later this year.

## Modernisation for East African Telephone Link

'I'EN years ago the Marconi Company installed a multichannel link between Nairobi, Mombasa, Tanga and Dar-es-Salaam in East Africa Now the East African Posts and Telecommunication and Administration has placed a contract with Marconi's to modernise the complete system which will result in improved performance and increased capacity for the telephone channels.
After modernisation there will be two radio paths over the whole link, carrying a total of 96 telephone channels, and automatic change-over facilities will ensure that priority traffic is not interrupted by failure of either radio path.

Despite the increased capacity of the links, the original terminal and repeater stations will he used. some of which link distances of up to 97 miles.

# A SIGNAL GENER ATOR 

# transistorised . . . a.f. and modulated r.f. outputs . . . battery operated . . . 

BY F. G. RAYER

THH1S signal generator uses two transistors, and will provide either an audio frequency output or a modulated radio frequency output. The a.f. output can be used to check audio amplifiers, or the a.f. stages of a receiver. The r.f. output is tunable from about $1.6 \mathrm{Mc} / \mathrm{s}$ to $170 \mathrm{kc} / \mathrm{s}$ (approximately 190 to 1.800 m ) and can be used to test or align intermediate frequency amplifiers, or medium and long wavebands of a receiver. Harmonics of $200 \mathrm{kc} / \mathrm{s}$ are audible to $10 \mathrm{mc} / \mathrm{s}$, while harmonics of $1 \mathrm{Mc} / \mathrm{s}$ are audible to $30 \mathrm{Mc} / \mathrm{s}$, with a sensitive receiver. This allows frequency spotting through the short wave ranges of a receiver.

The generator operates from a 4.5 V dry battery supply, and may be used for checking transistor sets, or mains or battery operated valve receivers.

The circuit is shown in Fig. 1. Trl is a radio frequency transistor, such as is commonly employed as a self-oscillating mixer in superhet receivers. Tr 2 is an audio type; a yellow-green spot audio transistor was used by the author, but any small transistor in working order should be satisfactory.

T1 is a transistor coupling or driver transformer, with a ratio of about $1: 2$ or $1: 1$. The transformer
fitted was intended for push-pull, so one secondary lead is unused. If $\operatorname{Tr} 2$ does not oscillate, connections to the primary of T1 should be reversed. With the switch S 1 in position 1, the audio signal is taken to the output control potentiometer VRI, which allows volume to be kept down as required.

Tr 1 is the r.f. oscillator. A two-gang 400 pF or similar value tuning capacitor is used to obtain adequate coverage with only two bands.

With the switch S 1 in position 2, the generator tunes from about $1.6 \mathrm{Mc} / \mathrm{s}$ to $500 \mathrm{kc} / \mathrm{s}(190$ to 600 m approximately). When the switch is in position 3, the coverage is approximately 500 to $170 \mathrm{kc} / \mathrm{s}$ ( 600 to 1.800 m approximately) and this includes most of the long wave band, in addition to $370 \mathrm{kc} / \mathrm{s}$ and adjacent intermediate frequencies.

C5 provides coupling from the a.f. oscillator, which acts as modulator, so that the note can be heard on a receiver tuned to the same frequency as the generator.

Small transistor type components can be used throughout, except for C6, which must be a mica or other high voltage capacitor, in order to provide effective isolation from mains equipment.


Fig. I: The complete circuit of the generator.

## Tuning Coil

This was wound as in Fig. 2, silk covered wire of $34 \mathrm{~s} . \mathrm{w} . g$. being used throughout. The m.w. section consists of 50 turns wound side by side, between points 1 and 2, a layer of paper being placed over the rod. A space of about $1 / 5 \mathrm{in}$. is left, and seven turns are wound side by side (leads 3 and 4). A narrow space (about $1 / 20 \mathrm{in}$.) is left, and 100 turns are wound in a compact pile, for the l.w. section. The end of the m.w. winding is joined to the beginning of the l.w. winding (lead 2). The end of the l.w. winding is lead 5. Dabs of cement are used to hold windings and the ends secure, and all turns must be in the same direction.


Fig. 2: Winding details of the tuning coil, LI.
In general, other coils should be satisfactory. However, some experiment may be needed. If coverage is too high in frequency flow in wavelength) additional turns are required; but if frequency coverage is too low, turns need removing. Air cored coils were found to work, but need many more turns. Some ready-made dual-wave coils could be expected to be satisfactory, though coverage might need modifying slightly, either by changing the core positions, or by altering the number of turns. With most coils, turns will need


Fig. 3 (above) and Fig. 4 (right): Component layout and wiring diagrams.
removing, to reach about $500 \mathrm{kc} / \mathrm{s}$ on the 1. . . bund, with VCl fully open.

## Construction

A paxolin panel about 4 in . x $4 \frac{3}{1} \mathrm{in}$. is used, drilled as in Fig. 4. The ferrite rod (L1) is a push fit in a $\frac{3}{8}$ in. diameter hole, being held with cement. A $\frac{1}{16} \mathrm{in}$. drill will do well for all the small holes, with $\frac{3}{8} \mathrm{in}$. hole for the switch and potentiometer and a $\frac{1}{2} \mathrm{in}$. hole to clear the gang capacitor.

The components are mounted as in Fig. 3. Emitter, base and collector leads of Tr1 and Tr2 are inserted as shown and there is no need to cut these wires short. The transformer T1 is held by its leads.

Three short 4B.A. bolts hold the gang capacitor. These must not penetrate sufficiently to shortcircuit or damage the capacitor. The circuit requires a three-way, two-pole switch and, if a surplus switch with more contacts is fitted, unwanted tags are ignored.

Fig. 4 shows wiring on the front of the panel between resistors and other parts. Any thin, insulated wire is satisfactory or thin tinned copper bare wire can be employed with 1 mm sleeving.

The audio oscillator stage can be tested alone when wired. To do this connect phones from C6 to battery positive. If no tone is heard reverse the primary leads to T 1 as mentioned.

Two short pieces of- thin flex are provided for battery connections and there was found to be no advantage in using more than 4.5 V . Consumption is between about 2 mA and 6 mA according to the actual transistors, so very small cells will do well. To avoid trouble from bad contact leads may be soldered to the cells. Three dry cells in series will provide 4.5 V and the zinc case is negative.

## Output Lead

A flexible lead from the positive line terminates in a clip which can be attached to the chassis or other earthed circuit of a receiver.


The output lead from C6 is fitted with an insulated prod so that the signal can be injected into various parts of a circuit.

There is no chance of shocks when testing a battery operated transistor receiver. With battery operated valve sets the h.t. may be sulficient to cause a shock. In the case of mains receivers the usual care must be taken. The danger is particularly great with a set of the a.c./d.c. type or one drawing h.t directly from the mains.

## The Case

The case is made of thin wood approximately $4 \frac{1}{4}$ in. $x ~ 4 \frac{3}{4}$ in. $x ~ 2 \frac{1}{4} \mathrm{in}$. deep internally. The panel is thin paxolin 4 in . $x 5 \frac{1}{4}$. if the case is of $\operatorname{tin}$. thick wood. Nuts on the switch and potentiometer secure the generator and panel together and scales are marked on the paxolin or on thin card placed over the paxolin. For tuning a fairly large knob is required and is fitted with a celluloid cursor marked with a line to permit reading against the scales.

To check the generator the a.f. output may be applied to a pair of phones or to an amplifier or receiver. To test for modulated r.f. output place the prod or lead near the aerial of a receiver. The audio tone should be heard when the receiver and generator are tuned to the same frequency.

## Harmonics

The generator note will also be heard when the receiver is tuned to a multiple of the generator frequency. For example. if the generator is tuned to $200 \mathrm{kc} / \mathrm{s}$ its note will be heard at $200.400,600$, $800.1,000.1,200 \mathrm{kc} / \mathrm{s}$ and so on. This fact is very useful both to calibrate the generator tuning scales and to furnish a signal for a short-wave receiver.

The correct tuning point will most easily be found when coupling between generator and receiver is low. and output small, so that the received signal is not too strong.

## Calibration

If a calibrated generator is available, set this to various frequencies, tune in the signal on a receiver, then tune the transistor generator to the same frequency and mark its scale.
If no calibrated generator is to hand, calibration can first be at $200 \mathrm{kc} / \mathrm{s}$ points. Find this fundamental point from the BBC Light Programme transmitter. Tune the generator to the same frequency and mark its long-wave scale. Then tune the receiver 10 a harmonic as mentioned. Leave the receiver tuning untouched and tune the generator until its fundamental is heard on the receiver. This can then be marked on the generator scale. Repeat the whole procedure until the scales are calibrated to $1.6 \mathrm{Mc} / \mathrm{s}(1,600 \mathrm{kc} / \mathrm{s})$. Other calibration marks can be oblained by selecting other harmonics. For example, the third harmonic of the Light Programme (long waves) is $600 \mathrm{kc} / \mathrm{s}$. The generator may then be tuned to $300 \mathrm{kc} / \mathrm{s}$ ( $600 \mathrm{kc} / \mathrm{s}$ is the second harmonic of $300 \mathrm{kc} / \mathrm{s}$ ) to obtain the $300 \mathrm{kc} / \mathrm{s}$ point between 200 and $400 \mathrm{ke} / \mathrm{s}$. Proceeding in this way, sufficient calibration points can be secured.

Note that the tone is not heard if the generator is tuned to a multiple of the receiver frequency but only when the receiver frequency is a multiple of

## COMPONENTS LIST

## Resistors:

| Resistors: |  | R4 | $27 \mathrm{k} \Omega$ |
| :--- | :--- | :--- | :--- |
| RI | $27 \mathrm{k} \Omega$ | R5 | $5.6 \mathrm{k} \Omega 2$ |
| R2 | $8.2 \mathrm{k} \Omega$ | R6 | $1 \mathrm{k} \Omega$ |
| R3 | $1 \mathrm{k} \Omega$ |  |  |

VRI $25 \mathrm{k} \Omega$ potentiometer with s.p. switch (S2)

## Capacitors:

$\begin{array}{ll}\mathrm{Cl} & 0.01 \mu \mathrm{~F} \text { paper } \\ \mathrm{C} 2 & 0.01 \mu \mathrm{~F} \text { paper }\end{array}$
C3 $0.01 \mu \mathrm{~F}$ paper
C4 50pF ceramic
C5 $\quad 0.01 \mu \mathrm{~F}$ paper
C6 $\quad 0.01 \mu \mathrm{~F}$ mica, or paper 500 V
C7 $0.01 \mu \mathrm{~F}$ paper
VCl $400+400 \mathrm{pF} 2$-gang variable

## Inductors:

LI H.F. coil (see text)
T1 Transistor coupling or driver transformer Switches:

Si 2-pole, 3-way miniature rotary
S2 s.p.s.t. (see VRI)
Transistors:
TrI OC44 or similar r.f. or mixer type
Tr2 OC72 or similar a.f. type

## Miscellaneous:

Two small pointer knobs. One $1 \frac{1}{2} \mathrm{in}$. diameter knob. Paxolin, wire, sleeving, clip and prod, etc.
Ferrite rod for L2 (2in. $\times \frac{3}{8} \mathrm{in}$.).
the generator frequency.
The signal generator may be used to test or align a receiver in the nornal way. The following brief details will help indicate how this work is carried out.

## Checking Audio Circuits

If a receiver or audio amplifier does not operate, a quick check of the audio stages can be made by utilising the a.f. output of the signal generator. An insulated test lead with prod is used to apply the a.f. output to various circuit points in turn, working backwards from the loudspeaker. When the point at which a fault arises is passed, reproduction will deteriorate or cease.
As an example the first test will show if the loudspeaker is working. The next test will indicate if the output transformer is in order and connected. The output valve or transistors are then brought into the portion of the circuit under test by injecting the signal at the control grid or base. The coupling capacitor or driver transformer can then be checked by removing the prod to the input side of this component. Earlier a.f. stages are then taken in ene by one.
This is a very rapid means of localising a fault. If such tests are carried out on a receiver or amplifier in good order this will show what to expect. When the a.f. signal is applied directly to a loudspeaker or outpit transformer and loudspeaker, volume is low. An earth return lead from generator to receiver chassis will be necessary when little or no amplification is available. To test a loudspeaker, output and earth leads must be taken to the speech coil. As one or more stages of amplification are introduced, by moving the output lead to grid or base, volume from the loudspeaker should increase considerably. The earth lead from generator to receiver is then often unnecessary.

If a fault is present in audio stages these tests will show where it arises. If the whole atudio section of a receiver works satisfactorily this shows the defect is in earlier stages.

## I.F. Testing, Alignment

If the receiver is already aligned, stage-by-stage tests of the intermediate frequency amplifier can be made in the same way as described for audio stages. However, the generator is tuned to provide a modulated radio frequency output corresponding to the receiver intermediate frequency. This is done by tuning the generator for best output from the receiver.

The signal may be injected at the final IFT primary. If the audio section has tested satisfactorily and there is no output from the receiver, then the IFT or detector or wiring here must be suspected. If output is satisfactory the prod is moved back to take in the last i.f. valve or transistor, then the preceding IFT and so on. The whole i.f. amplifier may thus be quickly checked.

The strength of the signal injected into the receiver is adjusted as required in a similar manner to that already explained. When a stage or two of i.f. amplification has been introduced into the part of the circuit being tested, loose coupling between prod and receiver circuit will generally be sufficient. That is, the prod point only need be placed near the receiver lead or against the insulation of the lead.
Should the receiver be newly built, and not aligned, the generator should be tuned to the wanted frequency (often $470 \mathrm{kc} / \mathrm{s}$ ). This signal is injected at the last IFT and the IFT core is adjusted for best output from the receiver. The previous 1FT is then taken in and adjusted, followed by any earlier IFTs present.

As capacity from the generator prod will slightly influence alignment, all the IFTs are given a final check to see if any slight readjustment is wanted. This is done with the generator coupled to the input (grid or base) of the mixer or frequency changer.

When i.f. or earlier stages are being tested the receiver volume control should be at maximum, but volume should be kept quite low by reducing the strength of the injected signal as described. This is to avoid the automatic volume control circuit of the receiver making accurate adjustment difficult.

If some form of output meter is preferred to adjustment by ear this is easily arranged. A 25 mA , 50 mA or other d.c. meter may be added in one battery lead of transistor sets having push-pull output. Alignment is then to secure maximum meter reading. An a.c. voltmeter with $0.5 \mu \mathrm{~F}$ or other blocking capacitor in one lead can be used as an audio output meter. A d.c. meter can be used to measure the anode current of an i.f. stage
to which a.v.c. is applied, adjustments being to obtain minimum current. The meter should be at the h.t. side of the IFT primary with a bypass capacitor to chassis. Alternatively a highresistance voltmeter may be used to measure the a.v.c. line voltage. Adjustments are directed towards securing an increased negative voltage with valve sets. With transistor receivers the a.v.c. line is positive.

## Mixer, R.F. Stages

Tests in r.f. stages are as described, the generator being tuned to the same frequency as


The instrument completely wired up, ready for installing in its case.
the receiver. To trim such stages adjust the generator to a high frequency (low wavelength). To make adjustments to coil or aerial inductance tune the generator to a low frequency (high wavelength). It is usual to deal with the m.w. band first, followed by long waves.

Sufficient input should be obtained by placing the generator prod or lead near the receiver aerial socket or frame aerial or ferrite rod. If trimming is initially badly out a fairly strong signal (closer coupling) may be used, followed by more accurate trimming on a weaker input (looser coupling).

To secure agreement with a printed dial, trimming and inductance adjustments are made as described for various frequencies after checking that the dial or pointer is properly fitted to the tuning capacitor. If a dial is to read accurately throughout, the tuning capacitors, trimmers, coils or aerial and padders must all be of the specified values intended for use together.

With mains operated equipment, remember to take the usual care to avoid shocks. Receivers of the a.c./d.c. type or sets drawing h.t. directly from the mains may have dangerous mains voltages present on the chassis, etc., if fed from a reversible two-pin plug or adaptor or from a wrongly wired three-pin plug.

# A Novel A.F. Amplifier 

BY E. A. PARKER

FUIG. 1 is the circuit diagram of an amplifier which has been used for some time past in conjunction with a simple regenerative transistor receiver to provide loudspeaker output for a fairly large living-room. In fact the output was sufficient at times to make the provision of a volume control essential. The amplifier was found to be capable of driving loudspeakers of diameters of from 3 in . to 8 in . with equal efficiency and has been incorporated into a small portablet.r.f. receiver using a 3 in . speaker.

## Directly-coupled Output Stage

The main design considerations were that the amplifier should be easy to construct, economical to run and should require as few components as possible-this saved space and money. It was
was the circuit of Fig. 1, which seems to provide a large current gain in addition to providing an output well in excess of 100 mW .

## Alternative Transistors

The transistors used were XB102s (or XB112s) for Tr 1 and Tr 2 and an XC101 (or XC 111 ) for Tr3, the output transistor.

Although Ediswan transistors were used the Mullard OC71 and OC72 (Tr3) are equally suitable. The amplifier also works well with surplus transistors (red spots for Tr 1 and Tr 2 and green/ yellow spots for Tr 3 ) but if these are used the directly-coupled pair should be selected so as to have low leakage currents.
The layout of the amplifier was not found to be critical and in view of this no constructional details are given, but Fig. 2 indicates the orientations of the connecting leads of the transistors. One point to note is that the emitter bias of the output pair tends to be slightly more critical than does that of a conventional single-ended output stage, but the value of $100 \Omega$ for the resistor R5 in the emitter leads as shown in Fig. 1 will be found to be satisfactory in most cases. Depending on the individual transistors used, the optimum value of this resistor should be found to be within the range $50-500 \Omega$.
The interstage transformer T1 , may be replaced by an R-C coupling circuit if economy demands this; the collector load

Fig. I: The circuit of the omplifier.
decided, therefore, that two of the transistors should be d.c. coupled. Now audio amplifiers capable of driving a loudspeaker are as often as not of the push-pull variety. requiring expensive push-pull driver and output transformers and matched transistors. So an attempt was made to construct a directly-coupled output stage.
As stated above, the final product of experiment



Fig. 2: Lead identification of types of transistors that con be used in this circuit.
should then be of the order of $4.7 \mathrm{k} \Omega$ and the coupling capacitor should have a value of $8 \mu \mathrm{~F}$ with a minimum of 12 V d.c. working voltage. A slight loss in stage gain will result with this arrangement.

## Output Transformer

The output transformer T2 shoild be chosen so as to reffect a collector load of about $48 \mathrm{k} \Omega$ to Tr 2 and $\mathrm{Tr}_{\mathrm{r}}$. If this load be designated by Rc and the

# The <br> Photopholif <br> By M. L. Michaelis <br> Experiments with <br> light waves - a new <br> interest for the amateur electronics enthusiast. 

## CONTINUED FROM PAGE 427 OF THE SEPTEMBER ISSUE

CONTINUING the discussion from last month on radiation, it may be said that the intensity of an electromagnetic radiation is given by the electric or magnetic field intensities on the wave picture and by the average number of photons arriving in a given time on the "photon" picture. It is necessary to stipulate "average" rate of arrival of photons in the second case because the actual sequence of arrival of photons, in common with all particle radiations, is quite random, subject to purely statistical fluctuations over short periods. This phenomenon has already been discussed in detail in connection with the digital Geiger counter apparatus published in this journal in the February, 1962, and following issues. The methods of interpreting quantitative results when forced to use the particle way of looking at things, rather than the wave method, at the very high frequencies represented by atomic radiations have also been described in the named articles.

Visible light represents just about the reasonable borderline case where it is about equally convenient to treat it statistically as a random sequence of particles, called photons, or to treat it as a continuous wave electromagnetic radiation. Indeed some optical devices are most conveniently treated on a basis of photons, whereas others are more conveniently explained on a wave basis, it always being necessary to remember that the two treatments amount to the same thing, and thus the practical engineer designing equipment operating at visible light frequencies will use the method of approach offering the greater simplicity in a given case.

A final point should bear-this out. We are normally accustomed to thinking of light as a wave radiation, which is indeed more often than not the most useful line of approach. However, the photophone unit described in this article is, an excellent example of the manifestation of the "photon" particle properties of light if we look more closely.

The detailed action is best represented by a "photon-for-electron" picture. Each photon particle of the incident light actively' reaching the cathode of the photocell ejects one electron contributing to the resulting anode current. Were the rate of arrival of photons absolutely steady, then the stream of electrons produced in the photocell would be quite steady for an unmodulated steady beam of incident light. An audio amplifier connected to P 2 would register no a.c. signal. Upon trying this in practice, however, we find that this condition is only approximately true. If the gain of the audio amplifier is very high, or at least reasonably high (e.g. both stages of an ECC83, followed by an EL84 output stage), a powerful hiss will be heard as soon as a steady beam of light is shone into the photophone. This shows that there are random statistical fluctuations taking place in the rate of arrival of the photons, leaving only a constant long-term average. Exactly the same hiss is heard if a Geiger counter is subjected to strong atomic radiation, giving a high average counting rate. The true average frequency of arrival of atomic particles, which may in that case be in the region of $10 \mathrm{kc} / \mathrm{s}$, is not heard primarily, but rather the band of frequencies represented by, the random variation, leading to a raucous hiss.

The same effect is found, yet again, in the, thermionic emission of electrons from a valve, cathode. The average rate of emission of electrons; representing the total saturation cathode current of: which the valve is capable, nevertheless contains: the short-period random statistical fluctuations; characteristic of all particle emissions and radia-; tions in Nature. These statistical fluctuations of: cathode emission give rise to the same characteristic hiss. responsible for the major portion of "valve noise" in sensitive input stages of amplifiers.

The same statistical laws anply for the absorption of particle radiation as for its emission, thus if several grids and anodes are receiving each a certain average portion of the total cathode current the exact momentary partition of current among the electrodes is subject to further random statistical fluctuation, giving further hiss. Thus, as will doubtless be familiar to many readers, multielectrode valves are more " noisy" than triodes.

## Use of Valve Voltmeter

With the valve voltmeter circuit of Fig. 4 it is seen that, when using the universal probe shown, connected to P5, d.c. ranges of 5,50 and 500 V f.s.d. are available.

The upper range is not useful for the Photophone, being intended for general usage of the valve voltmeter, but the lower two ranges are so intended. For this purpose P5 of the valve voltmeter is connected directly with P1 of the Photophone without any probe. The probe


Fig. 4: A high quality valve-voltmeter circuit for use with the Photophone.
resistor R38 is missing under such circumstances so that the effective sensitivity is doubled, f.s.d. being obtained for 2.5 and 25 V input respectively.

As the voltage calibrations on the meter for correspondence to readings with the universal probe are 5 V or 50 V f.s.d., and we have seen that an'output of 0.5 V per foot-candle is to be expected, the valve voltmeter readings on the existing voltage scales give foot-candles directly.

If, upon more accurate comparisons, corrections are found necessary to bring the voltage calibrations of the meter in Fig. 4 and the foot-candle values into true correspondence, various measures can be adopted. Within limits the values of R5 and/or R6 in the Photophone may be adjusted if only small corrections are required either way. If larger reductions are found to be needed, then suitable resistors may be included in series in the cable connecting P1 and P5.

Larger increases of sensitivity are best obtained by fitting a switch to select different values for R 25 and R26 in the valve voltmeter when switching to "photometer" setting on this switch. Much greater overall sensitivity is easily obtainable by using a more sensitive meter. This can be fitted with a universal shunt for switching back to the 2 mA f.s.d. sensitivity for normal use in the design of Fig. 4 or for multiplying the sensitivities of the incorporated voltage scales.

## The Valve Voltmeter

It is not proposed to go too deeply into the principles of the circuit but merely to give sufficient information to enable the reasonably experienced reader to build the valve voltmeter.
The design is based on a simpler version which was published in the December, 1961, issue of Practical Wireless. This earlier design suffered to some degree in the matter of obtaining linearity up to full-scale deflection.
However, some further experiments have in the meantime been undertaken on this circuit, Fig. 4 showing the result. For the price of some appreciable increases in complexity, it was ultimately found possible to get good linearity to full-scale deflection for both positive and negative inputs, polarity-switching being undertaken by changing over between the two grids of V4 by means of S 3 .

The same 2 mA meter was able to be retained and in fact virtually the same layout and design as published in December, 1961. Thus the same disposition of tagstrips was maintained and those readers who have already built the simpler instrument could very easily and quickly convert to the new comprehensive form.

Those building the unit new from scratch will find that the layout is not critical except for the need to mount the grid stoppers R27 and R29 very
close to the grids concerned and to connect nothing else directly to the grids. These grid stoppers were not required in the simpler circuit because this operated at much. lower standing anode current, where parasitic oscillations were far less likely than in the present circuit.

In all other respects meticulous insulation, to avoid errors due to leakage in the high-impedance circuits, is far more important than any particular layout which may be chosen to suit individual requirements and available materials.

## Methods of Cathode Bias

To get the present circuit to operate on negative inputs as well as positive ones, the operating point of the valve had to be shifted to higher current, at the same time retaining the relatively high cathode load to give proper coupling between the two sections. The only way of achieving this was to bias the cathodes down to a negative voltage, which was conveniently obtained at sufficient current capabilities by means of D5 and C7 from the heater line.


Fig. 5: An a.c. probe usable throughout audio, long wave, medium wave and short wave ranges, and able to tolerate all voltages present in normal receiver and amplifier circuitry.
(See also table below).

$$
\begin{array}{ll}
\text { Full scale deflection at } & 50 \mathrm{c} / \mathrm{s} \text { mains frequency } \\
\text { input, using this probe in } \mathrm{P5} \text { of the prototype valve } \\
\text { voltmeter (Fig. 4) represents.- } \\
\text { VTVM switched to: } & \text { Sensitivities: } \\
50 \mathrm{~V} \text { r.m.s. a.c. range } & \text { Do not use here } \\
5 \mathrm{~V} \text { d.c. range } & 15 \mathrm{~V} \text { r.m.s. } \\
50 \mathrm{Vd.c.} \text { range } & 150 \mathrm{~V} \text { r.m.s. } \\
500 \mathrm{~V} \text { d.c. range } & 500 \mathrm{~V} \text { r.m.s. }
\end{array}
$$

Notes: (1) The sensitivities given hold true up to about $100 \mathrm{kc} / \mathrm{s}$ and down by factor six at $20 \mathrm{Mc} / \mathrm{s}$ in the prototype. (2) On the 500 V d.c. range do not exceed 250 V r.m.s. input.

This, incidentally, is a most useful general method of biasing a cathode negative wherever this may be required in a circuit to enable the cathode-load resistor to be increased without choking back anode current. The converse of biasing a cathode more positive still than the voltage given anyway by virtue of anode current through the cathode load, is available by means of taking an additional high resistor from the cathode to h.t. positive, a measure capable of driving the valve even beyond cut-off in the resting state. This is, of course, never possible with a cathode load alone, for a cut-off valve takes no anode current and thus cannot produce bias across a cathode load, and thus could not be cut off, which contradiction proves the impossibility.

## Trimming the Valve Voltmeter

The combination R31, R32, as in the earlier circuit, remain the crucial components for proper operation. Too low a value gives insufficient deflection sensitivity on the meter, too high a value severe non-linearity in the form of cramping near full scale. This might, however, be considered useful in the interests of opening up low readings and those readers taking this point of view could use an appropriately calibrated non-linear scale attached to the meter.

To reduce thè actual current swing required in the valve for f.s.d. meter defiection, it is useful to ascertain that as little current as possible is normally shunted through VR3. For this purpose R25 and R26 are simultaneously adjusted in such

a manner that the dividing ratio is altered, bat the sum of R25 and R26 in parallel with 1.M 2 remains equal to $1 \mathrm{M} \Omega$-in order that the input voltage dividers to the left of $\$ 2 \mathrm{~b}$ remain correct.
This adjustment of R25 and R26 is undertaken on the "calibrate" setting, with VR3 disconnected, until the meter only still reads very slighty high, final correction then being continuously available on VR3 when this is reconnected. This adjustment of R25 and R26 has already been carefully undertaken in arriving at the component values specified and these should be found satisfactory.

If trouble is experienced. readers should first check for low emission in V4 by substitution and also check that R25 and R26 are of the values specified.

## Further New Features

While modifying the originat simple valve voltmeter circuit io a form ideal for use with the Photophone, other changes were made to improve its general usefulness, again at some additional expense and complication.
The two grid circuits have been fully symmetrised and kept to normal impedance levels of around $1 \mathrm{M} \Omega$. This avoids trouble from zero point drifts due to electron impact grid current otherwise present in high-impedance grid circuits and also makes the zero point fall at virtually the same setting of VR2 on all ranges.

Furthermore the zero point now falls very close to the middle position of VR2 under all conditions and requires only very infrequent adjustment after an initial five to ten minutes' warming up. These advantages have been obtained at the price of one disadvantage over the original., namely a reduction of overall impedance presented as load across the voltage source being measured. However, a reasonable compromise has been struck.

The disposition of smoothing capacitors has been changed to give a quicker response and quicker return to zero and a high resistance has been removed to the probe head to prevent a.c. damping stopping, for example, oscillators functioning when d.c. potentials at electrodes are to be measured.

Furthermore the whole aggregate around DI and D2 has been incorporated to give a 50 V r.m.s. f.s.d. range for a.c., usable at mains frequency only, without changing probes. The combination R17 to R19 should be adjusted until this scale reads correctly; it may be possible to select a single $56 \mathrm{M}!2$ resistor in place of these three. The new circuit uses a pair of diodes, D1 and D2, for achieving high inverse voltage rating, preventing damage even if the a.c. signal is superimposed on a d.c. voltage of some hundreds of volts, which is blocked at C 5 .

It is thus possible to measure the gain of amplifier stages with this instrument, using a $50 \mathrm{c} / \mathrm{s}$ sinewave signal and comparing levels at grid and anode, without any need to be concerned with d.c. potentials which may or may not be present. Finally the work on this valve voltmeter was completed with the design of a special a.c. probe, for use at other frequencies, operating into the d.c. ranges. This rectifies at the probe head in the usual way so that high-frequency signals do not have to pass down the coaxial cable and get lost in the capacity thereof.

Experiment revealed the interesting fact that, although D6 is a silicon rectifier intended for mains power rectification, usable performance was possible at frequencies up to $20 \mathrm{Mc} / \mathrm{s}$. The high inverse voltage rating of the silicon dlode is a great advantage, so that this probe cannot be damaged by any normal h.t. voltages found in receivers and amplifiers. Otherwise it is of the conventional form already mentioned in the other publications.

## Conclusion

It is hoped that this article has been successful in introducing the possibilities of optical frequencies for radio experiments as well as explaining some of the significances and aspects of various types of radiation in the light of modern science in a form which is surely of a more practical usefulness for the understanding of one's equipment than mere philosophical theory for which it might at first sight be mistaken.

## Novel A.F. Amplifier

-continued from page 500
loudspeaker impedance by RL, then use a transformer of turns ratio $n=$ number of primary furns number of secondary turns where $n^{3}=R c$

> RL

The following table gives the value of $n$ for three popular loudspeaker impedances. (Here, Rc $4.7 \mathrm{k} \Omega$.)
Loudspeaker impedance

| (RL $\Omega$ ) |
| :---: |
| 8 |
| 3 |
| 35 |


| Output transformer <br> turns ratio $(\mathrm{n}: 1)$ |
| :---: |
| $40: 1$ |
| $25: 1$ |
| $10: 1$ |

The transformer used by the writer was abstracted from a defunct battery operated valve portable receiver-its physical size was small enough.

## Additional Pre-amplifying Stages

Clearly the number of transistors preceding the output pair may be varied according to the number available and the final output required; three such transistors would overioad the power output transistor Tr 3 on loud passages of music if the amplifier is used with an average local station t.r.f. receiver. Only one was required in the writer's home town (Orpington, Kent) to provide ampie listening volume on the Light Programme ( 247 m ) and the Home Service $(330 \mathrm{~m})$ from a one-transistor regenerative receiver with a 5 in . loudspeaker connected to the amplifier outputs.
The amplifier is powered by a PP7 ( 9 V ) battery which has a life of more than 70 hours when the amplifier is run at full volume.

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 justified with equipment required for military or similar purposes, they are not warranted (or indeed desirable) with more commonplace equipment. But, let me emphasise, guard against going to the other extreme of merely bringing the two surfaces into contact and relying on a generous blob of solder to do the rest. True, I admit, that this procedure can be quite effective with some miniature components where mechanical strain is negligible. Even so, I consider it advisable to make a self-supporting connection wherever this is at all possible. This is certainly no hard task, and the following suggested methods will usually suffice.

Push the lead through the hole in the tag and, if it is a loose fit, bend the protruding portion just sufficiently to provide a locking effect, then cut off the surplus wire fairly close to the tag. With larger components it is sometimes necessary to make a full right-angled bend and to squeeze the end of the wire against the flat surface of the tag. Where the lead is not passed through a hole in the anchoring device, it should be hooked around the edge of the tag or terminal.

## A Tape Teaser

One of the interesting uses to which magnetic tape recording is now being applied is the training of foreign language students in correct pronunciation. Recording systems known as "Language Laboratories" allow a number of students to perform oral exercises simultaneously without causing interference.

The usual method of operation is like this: each student sits in a booth equipped with a special tape recorder and, via headphones, listens to speech pre-recorded by the tutor. This recording is divided into a number of fairly short passages, and each is followed by a pause of similar duration. At the end of each passage the student repeats this phrase or sentence into his microphone endeavouring to copy faithfully the pronunciation of the original.

Whenever he wishes, the student can rewind his recorder and set it to playback. He then hears once again the tutor's voice followed immediately by his own efforts at imitation. For further practice he resets to record and proceeds as before, his earlier efforts being erased in the process, as is normal.

The little problem I now leave you with is this: how is it contrived that the tutor's recording remains preserved throughout and cannot be erased by the student no matter how often, he re-records his own voice in the gaps provided?

# PERSONALN <br> MTRANSISTOR TWO 

A Simple Medium Wave Set Suitable for the Beginner

BYA. SYDENHAM

0NLY a comparatively small cash outlay is involved in the construction of the receiver to be described and little or no test equip* ment is required to bring it into use.

Quite often a newly constructed superhet fails to provide any kind of signal when first switched on due to severe misalignment, but this is not likely to occur here as only a single tuned circuit is used. This set can, if necessary, be built quite easily on the kitchen table and no risk of lethal mains shocks is likely since transistors are used, two being employed since this is about the minimum number capable of providing a decent signal in a pair of headphones.

The prototype has also been tested with a loudspeaker by substituting the primary of a ratio 8:1 transistor out put transformer in place. of the headphones and connecting a $3 \Omega$ loudspeaker unit to the secondary winding. If the surroundings are quiet, as in a bedroom, for example, programmes can be heard quite well.

The receiver is powered by a single 4.5 V dry battery and the current consumption is so low (less than $\operatorname{lmA}$ ) that replacement is likely but rarely.

The frequency coverage is 600$1460 \mathrm{kc} / \mathrm{s}(500-205 \mathrm{~m})$, this covering practically the whole of the medium waveband. If longwave reception is required this can be incorporated but has not been attempted in the prototype.

In common with all simple t.r.f. circuits arranged to work at their most sensitive point the receiver is likely to prove slightly temperamental at the first try-out and slight adjustments to certain of the components might be occasioned. This point has been taken care of in the layout, however, which if followed accurately will allow changes to be made very easily. The range of component tolerances and the pro-
duction spreads in the transistors, etc., cause discrepancies that cannot be disposed of cheaply.

## The Circuit

In Fig. 1 it may be seen that the first transistor Trl is operated as a regenerative demodulator feeding a stage of audio amplification built around Tr2, high impedance headphones being inserted in the collector circuit of this transistor to enable the user to hear the selected programme.

The circuitry around Trl might be more easily understood by referring to Fig. 2, where a typical

Fig. 1: The circuit.

valve oscillator circuit is shown. Here the external tuning of the coil L is effected by capacitors Cx and Cy . Sometimes no resistor is included in the cathode circuit, bias for the valve being derived entirely from grid current. This circuit is widely used, the triode usually being part of a frequency changer valve when $C x$ is often made one section of a twin-gang tuning capacitor with Cy operating as a "padding" capacitance.


Fig. 2: A typical valve oscillator circuit.

Looking now at the circuitry around Trl (Fig. 1) the similarity will immediately be apparent and it will be appreciated that the transistor will oscillate if operating conditions are suitable.

By controlling the emitter current by means of a variable resistor bypassed by a capacitor, detection can occur if the transistor is not allowed to oscillate but instead is brought close to the point of the occurrence.

A suitable impedance match into the base of Tr can be effected by employing a capacitive potentiometer across the coil and using one section of it for tuning. Capacitors VC1 and C3 perform

this function and are in effect identifiable with Cx and Cy respectively in Fig. 2.

Since these capacitors are connected across the coil in series the total value of capacitance available will always be less than that of the smaller of the two. This reduced capacitance effect will make itself most apparent at the low-frequency end of the band, i.e. where the vanes of Cl are most fully engaged, and if a suitable coil is not used for L1, only a section of the band will be receivable. Fortunately suitable coils are available from the Denco miniature range.

A miniature Jackson "00" twin gang with sections paralleled ( $208+176 \mathrm{pF}$ ) form a total value of 384 pF , ignoring the trimmers, but this is reduced to about 330 pF by C3 and is then suitable for use with the coil specified. When another coil designed to work with a 500 pF tuning capacitor was tried in the circuit maximum low frequency coverage was reduced to $750 \mathrm{kc} / \mathrm{s}$.

Note that two capacitors- -C 1 and C 2 -are fitted between Trl collector and L1; this is not a safety


Fig. 3: The modified circuitry necessary to provide long wave coveroge.
measure but an aid to obtaining good results on completion, since these items control the amount of positive feedback to a great extent.
The efficiency of the receiver largely depends upon the characteristics of $\mathrm{T}_{\mathrm{r}} 1$ and this should be a good specimen; the use of a surplus type is not recommended.
The small r.f. choke of $3.000 \mu \mathrm{H}$ inductance (L2) fitted in the collector circuit of Tr1, consists of the main winding of a dust-cored long-wave coil, the primary winding being removed and the coil former sawn off to the length of the core.

As mentioned earlier, inclusion of long waveband facilities can be provided either by fitting a separate coil from the Denco range (Range No. 1) or by using a miniature dual-waveband coil as shown in Fig. 3. The additional switching required can be incorporated with S1 by using a rotary switch. A two-pole, three-way switch used as shown will allow the central position to be "of "


- Fig. 4: Dimensions of the control support panel. with "medium" and "long" situated on either side. This makes accidentally leaving the receiver switched on less likely than in other arrangements. No extra panel control will be needed if this idea is followed.


## Constructional Notes

The receiver is built around a miniature 18 -way tag board to which is affixed a simple metal panel that carries the three controls. This assembly is finally located on a main hardboard panel to which simple sections are finally fitted to form a cabinet. A short throw-out type aerial is required and this, together with the headphones, is plugged into sockets provided on one side. The completed receiver is $7 \frac{3}{4} \mathrm{in}$. wide and 5 in . high and the front slopes backwards slightly, for the sides taper from 3 in. at the base to 2 in. at the top.

Preliminary work consists of constructing the metal control support panel from 16s.w.g. aluminium or tinplate and details of this are given in Fig. 4. Prior to bolting this panel to the tag-board, one of the small holes running down the centre of the board must be enlarged with a circular file to $\frac{1}{4}$ in. diameter-as indicated in Fig. 6-to accept the coil mounting stem. The coil is locked in position by means of a polystyrene nut provided and this should only be twisted thumb tight, if pliers are used the stem will fracture.

## Wiring Up

The remaining construction consists of mounting and wiring the various components to agree with the above and below plans shown in Figs. 5 and 6 respectively, leaving the transistors to one side until a good check has been made.

When satisfied that no mistakes have been made the leadout wires of the transistors may be sleeved and then passed through existing holes from above the board. These may then be soldered to the tags shown, using a heat shunt. The actual transistor shells should


Fig. 5: Showing connections between the tag board and control panel. The tagboard has been drawn in the same plane as the panel for clarity.


Fig. 6: The reverse side of the tag board.
the VR1 position, but care should be taken not to advance the control excessively or the programme material will be supplanted by undesirable oscillation.

## Final Adjustments

The next step lies in adjusting the waveband coverage. If no signal generator is available transmissions must be used and the various regional transmitters can most easily be recognised at such times as separate news bulletins are being radiated. After dark Radio Luxembourg might be heard when the vanes of VCl are almost fully disengaged. The coverage can be varied within limits by adjusting the core of L1 by means of the brass stem fitted and also by changing the trimmer positions.

## Foults

If the receiver fails to oscillate at any setting of VR1 or VC1, either C1 or C2-hul not hoth together-may be temporarily short-circuited to increase feedback. Resistor Ri might need changing also if an indifferent transistor is in use and the value may be reduced to about $56 \mathrm{k} \Omega$. Noise may result in the form of a rushing sound if R1 is excessively reduced in value and $39 \mathrm{k} \Omega$ is about the limit.

If on the other hand it is not possible to eliminate oscillations at any setting of VC1 or VR1, remove C2 and replace it with one of smaller value -say 10 pF -and try again. Emitter resistor R4 can also be increased to, say, $680 \Omega$ or even $820 \Omega$,
this further limiting Tr1 current. The receiver should, of course, be switched off to make alterations.

## Cabinet

Since negligible weight is involved it is possible to construct a simple cabinet from hardboard using a frame of $\frac{3}{5}$ in. quadrant beading mitred at the corners to improve the frontal appearance. The dimensions of the case used for the prototype are shown in Fig. 7 the whole being held together with impact adhesive and small reinforcement strips.

With the aerial and phone sockets mounted on


A rear view of the receiver nearing completion.


Fig. 7: The canstruction of a simple cabinet for the receiver.
one side, ample space exists for the battery which can if necessary lie flat on the cabinet base plate beneath the tag-board or stand at the rear. On completion the case may be lacquered or otherwise finished to suit individual tastes.

## Conclus'on

Although this simple little set can provide a great deal of constructional and listening enjoyment, it obviously has its limitations and cannot be expected to compare with the high sensitivity and selectivity of the superhet form, for, of course, it is not in the same price class.

The receiver can be used as a simole tuner to feed an existing audio amplifier by connecting a $4.7 \mathrm{k} \Omega$ resistor in place of the phones and taking the audio from between Tr2 collector (via a $0 \cdot 1 \mu \mathrm{~F}$ capacitor) and the positive line. No such comuection must be made under any circumstances to d.c. apparams or to so called a.c. cquipment that derives its h.f. direct from the mains supply or danner to life will result.

# The EUPHON <br> <br> A frequency seloctive <br> <br> A frequency seloctive unit for feeding two unit for feeding two loudspeakers 

 loudspeakers}

## By I. Kampel

THE unit about to be described is a simple and inexpensive means of obtaining a spacial effect in audio reproduction. With the addition of the two loudspeakers required, the total cost of the author's model was about two guineas only, and this included the cost of making the walnut cabinet.

Two loudspeakers are set up as for stereo, but instead of two channels, one loudspeaker handles the lower frequencies, while the other handles the higher frcquencies. With the aid of the balance control an effect similar to stereo is achieved as the


Fig. I: The simple circuit of the Euphon.
higher pitched instruments come through on the tweeter and the bass instruments on the woofer.

The idea is based on the familiar divider network, used in modern audio equipment, but the original crossover circuit is not suitable when the speakers are separated as there will not be enough volume in the tweeter to give the required effect. The conventional divider network uses paper capacitors and extra components, whereas the Euphon uses ordinary electrolytics.

Fig. 1 shows the circuit. On the actual model there are single-type wander-plug sockets at the back of the cabinet, to plug in the bass and treble loudspeakers, and also sockets for the radio input. Numbers 1 to 6 indicate these sockets.

The circuit incorporates four identical electrolyctic capacitors coupled in pairs, can to can, this


Fig. 2: An end view of the cabinet.
arrangement having the same effect as a single paper capacitor, though not the same value as used in the conventional divider network. The two coils are the most expensive items ( 7 s .6 d . each) and are of $3 \Omega$ to match the output transformer. The potentiometer of $25 \Omega$ simply acts as a balance control between the two loudspeakers to obtain bass or treble emphasis, or a perfect balance as required.

## Cabinet construction

The whole unit may be built into a cabinet $7 \frac{1}{2}$ in. $x 4 \frac{1}{4} \mathrm{in}$. $x 2 \frac{1}{2} \mathrm{in}$. A 3 mall square of fin. five-ply will be good enough for the first stages of construction. Fig. 2 shows an end elevation. The cross shaded triangle in Fig. 4 is a small length of $\frac{1}{2}$ in. quadrant. The two end pieces of Fig. 2 fit on to the sides of the $6 \frac{3}{3} \mathrm{in}$. x $4 \frac{1}{4} \mathrm{in}$. baseboard. However, the top, which is made to slide out, rests on the top of these two edges and hence is $\frac{1}{2} \mathrm{in}$. greater in


Fig. 3: This front elevation shows the wooden runners fitted to the sides.


Fig. 4: How the top is slid into place along the runners and tops of the sides.
length. Fig. 3 shows this, the shaded portions being the ends.
Fig. 4 shows the method of sliding the top into place, the back being permanently joined. Crossed shading indicates the extra shaped pieces of ply fitted to the top panel to run next to the walls (also indicated by crossed shading in Fig. 3). The piece of oblong ply cut from the wood slider is fixed on to the inner wall of the cabinet to form the runner (see Fig. 4). Panel pins and glue will hold this construction firmly; however, the front panel should not be affixed yet, but be made to slide out freely.


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SUPAMOLD (a paper dielectric) with voltage ranges up to 600 v. D.C. at $85^{\circ} \mathrm{C}$ or 750 v. D.C. at $70^{\circ} \mathrm{C}$, and having an insutation resistance at $20^{\circ} \mathrm{C}$-after one minute-of 10,000 megohms or $2,000 \mathrm{ohm}$ farad, whichever is the less.
Power factor for both Duomolds and Supamolds is less than .01 at $1 \mathrm{Ke} / \mathrm{s}$ at $20^{\circ} \mathrm{C}$.
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Fig. 5 (left): Showing the use of Formico to provide a suitable finish to the cabinet.
Fig. 6 (right): The three double wander-plug holders at the back of the cobinet.

To make a professional job of the cabinet Formica panels can be cut to cover the whole of the cabinet, excepting the base. Most "do-ityourself" shops have a selection of cheap offcuts, and the author found in this way an offcut which had the appearance of expensively polished walnut. This costs only is.

Fit the Formica panels to the plywood case, seeing that the Formica sides are a little over tin. above the ply sides, so that these meet the Formica panel on the sliding lid. All edges of Formica should be bevelled (Fig. 5), so that they fit snugly together. Take care not to stick the lid in when fixing these. Also see that a length of Formica is fitted over the quadrant to be at top of front panel.

The frame round the front panel may be painted black or surrounded with Formica. The

| COMPONENT LIST |  |  |  |
| :--- | :--- | :---: | :---: |
| LI, L2 | Divider network coils, $3 \Omega$ |  |  |
| (Denco) |  |  |  |
| CI-C4 | $32 \mu \mathrm{~F}, 450 \mathrm{~V}$ |  |  |
| SI | TOggle switch |  |  |
| Sockets A-C Double type mounted on small |  |  |  |
| VRI paxolin strip | $25 \Omega$ potentiometer |  |  |

LI, L2 Divider network coils, $3 \Omega$ (Denco)
Cl-C4 $32 \mu \mathrm{~F}, 450 \mathrm{~V}$
SI Toggle switch

VRI $25 \Omega$ potentiometer
latter may be a little tricky however as allowance has to be made for the sliding lid. Fig. 6 shows the back of the cabinet and the three double wanderplug holders, which can be obtained on a small strip of Paxolin.

Components should be arranged in the cabinet after the potentiometer and toggle switch have. been mounted on the front panel, and the front panel secured. As it is advisable to mount the coils at right angles, a small strip of ply will hold one coil in place leaving space for sliders to pass, whilst the other may be mounted straight on the reverse side of the front panel.

As the input and output sockets are mounted on the back which comes out with the lid, the wires going to these should be left a little (not too much) long to allow for this.

## The Euphon in use

A certain amount of care is necessary in the choice of loudspeakers. The h.f. loudspeaker should be small ( $2 \frac{1}{2} \mathrm{in} .-6 \mathrm{in}$.) and the l.f. one larger ( $8 \mathrm{in} .-12 \mathrm{in}$.), both being matched at $3 \Omega$. The author uses a 3 in . and a 7 in . with excellent results.

However, the most important point to stress is that it is absolutely essential to have a good bass cabinet. If the l.f. loudspeaker does not have this, results will most certainly be disappointing. Orly a small cabinet is necessary for the tweeter, which may, of course, be fixed in a cabinet to match the woofer.

Set up the loudspeakers as for stereo, and sit between them adjusting balance. This system is mainly for music and has no application on speech only. Loudspeakers should be 6 ft to 12 ft apart, and connections should be reversed on one loudspeaker until, by judgment, the best arrangement is found.

It should be noted that when used to simply improve the reponse of an old radio, it might have very little effect. It will only reproduce the frequency ranges supplied to it, and if the receiver cuts off below $10 \mathrm{kc} / \mathrm{s}$, or before it reaches the higher notes, nothing will be gained by the use of this unit.

# A Versatile <br> DOUBLE-TRACE OSCILLOSCOPE 

## By J. H. B. Gould

This series deals with the construction of instruments based on the 5in. and bin. cathode-ray tubes currently available on the surplus market. The series of articles cover the basic instrument and progressively more advanced versions that incorporate a timebase, a.c. amplifiers, a square-wave generator and doubletrace facilities. The individual constructor can decide for himself how far to go-according to his inclination and requirements. A further article will deal with the use of the instrument as an oscillograph. This month, the basic instrument described in the September issue, is taken two stages further.

## CONTINUED FROM PAGE 46I OF THE SEPTEMBER ISSUE

PROBABLY the most useful piece of auxiliary equipment to be found in any normal oscilloscope is the timebase or sweep generator. This makes it possible to analyse a waveform on a linear time scale. The timebase details which follow enable the basic oscilloscope described last month to be extended to Instrument Type 2.

## The Generator

The generator described here (see Fig. 10) consists of one valve (V5) connected as a "standard" Miller-integrator with transitron maintaining oscillator. As its output is not great enough to drive the spot the full width of the screen without assistance, an amplifier (V6) is used.
The amplifier is well above its job so that, at high gain, it provides sufficient output for "expanded timebase" working; this makes it possible to enlarge a small part of the trace, chosen by the shift control, for closer examination: The load resistor of V5 takes the form of a potentiometer VR6 that serves as a "Sweep Amplitude" control.

## Synchronisation

In practice it is impossible to freeze a waveform on the screen using a free-running timebase; it will always tend to slip one way or the other. To prevent this, part of the waveform can be taken, amplified and applied to the transitron oscillator, causing the timebase scan to lock on to the obseryed waveform. This is the function of V4, the sync amplifier.

In some circumstances it is desirable to lock the timebase on some other frequency (in frequencycompatison tests, for example), so the "sync" input comes from a panel socket. To save using a switch here, the internal synchronising supply
from the " $Y$ " input is brought to this socket by a wander plug that can be transferred to a dummy socket when an external synchronising waveform is being used.
To avoid distorting the observed waveform, the minimum possible synchronising input must be used, so V4 is also equipped with a gain control, "Sweep sync" (VR5).

## Timebase Frequency

With this circuit, the sweep frequency can be varied from about $10 \mathrm{c} / \mathrm{s}$ to about $30 \mathrm{kc} / \mathrm{s}$ in six bands. Switching between bands is carried out with the aid of the "Y Selector" S3. In addition to selecting the band-setting capacitors for the


Fig. 9: The h.t. power supply for the timebase valves.


Fig. 10: Timebase generator, timebase and sync amplifiers.
sweep generator, S 3 also decides whether the $X$ plates will be fed from the timebase or from sonie, external source via the "X" terminals on the panel.

For this reason, the switch has two wafers, the wiring to the second of these being given in Fig. 11. (If it is intended to build amplifiers into the oscilloscope, $\mathbf{S 3}$ will need three wafers; the wiring to the third is shown in Fig. 13.


Fig. II: Selector switching (without amplifier).

Fine control of the sweep frequency is made pussible by potentiometer VR7. Most people will find it easier to set up a waveform with this control if the slider moves positive as the control knob is turned clockwise.

## Power Supplies

A power pack is necessary to supply the timebase valves and a suitable circuit is suggested in Fig. 9. This power unit includes a circuit for providing a 50 V peak-to-peak output for the calibration terminal.

## Deflector Polarity

In the timebase generator, the working part of the sawtooth waveform is negative-going with

Left-The outhor's finished model.

respect to time. The waveform is then inverted by the amplifier before being fed to the plates.

In order to provide a timebase crossing the screen from left to right, a positive signal at the "X" terminal should deflect the spot to the right, i.e. precisely as described for the simple oscilloscope.

## Sweep Generator Output

As will be seen from Fig. 11. the output of the timebase amplifier goes to a panel terminal. If a wobbulator is to be used with the oscilloscope, the sweep-frequency drive will be obtained from this terminal and not from " $X$ a.c." which leads nowhere when the timebase is working.

## INSTRUMENT TYPE 3

By adding a.c. amplifiers and associated switching, the development of the progressive oscilloscope can be taken a stage further. The details which follow, discuss the progression from Instrument type 2 to Instrument type 3.

## Amplifiers

"A1" and "A2" are wideband a.c. amplifiers, identical to each other, each consisting of a triode voltage amplifier and a cathode follower. The circuit is given in Fig. 12. As double triodes are used in the circuit, two valves only are necessary.


Fig. 12: The circuit of the Al amplifier-the A2 amplifier is identical.


## FOR INSTRUMENT TYPE 3

(a.c. amplifier version Figs. 12 and I3)

Resistors:

All $\frac{1}{2} W, 10 \%$ except where stated.
Capacitors:

All 0 K
TCI 3-30pF trimmer
Valve:
V7 6SN7
Switches:
\$3 3-pole 8-way wafer type rotary
$\$ 4$ 3-pole 4 -way if Instrument type 3 only is to be built. 4 -pole 5 -way if Instrument Type 4 is to be built

The only controls are the two gain potentiometers.

The amplifiers should not be used at low levels of gain if highfrequency components are present in the signal, as the valve input capacitance will act with the potentiometer as a frequencyselective filter, distorting the waveform before it appears on the screen.

## Switching Scheme

Fig. 13 shows the switching scheme for use with these amplifiers. Both the switches are of the rotary (wafer) type.


The first pole of the $X$ selector switch " $\$ 3$ " is used in the timebase generator to select the band-setting capacitors. The first six positions are concerned with the timebase and the seventh connects the input terminal directly to the deflector plate; only in the eighth position does the "A2" amplifier come into play on the $X$ plates.

The input signal should be fed to the "X a.c." or "X d.c." terminals regardless of whether the selector switch is at "Plates" or "A2"; this saves a lot of trouble in swopping-about wires.

The "Y Selector" switch " $\mathbf{S} 4$ " offers a choice of four ways in which a signal can reach the plates from the $Y$ terminals: attenuated, direct, amplified by one amplifier or amplified by both amplifiers in cascade.

## The Attenuator

For good high-frequency performance, the trimmer capacitor shunting the upper arm of the attenuator should be set to balance out stray circuit capacitances. A square wave applied to the Y terminals will soon show up any, faults in the frequency characteristic of this circuit.



Fig. I: Two-volve t.r.f.

WHILE a v.h.f. tuner in most areas will provide the finest local radio input for high fidelity equipment, the quality of some medium-wave broadcasts justifies the addition of a small a.m. unit. Leaving local transmissions to the existing tuner, the new unit can be designed for a balance between sensitivity and fidelity to make the best of those Continental stations that reach the aerial relatively unblemished.

Superhet selectivity is undesirable because of bandwidth considerations; on the other hand the usual fidelity t.r.f. designs are too insensitive for any distant signals. The unit to be described, however, has a high sensitivity for a two-valve circuit and brings in the worthwhile transmissions with satisfactory quality.

## The Circuit

The circuit, Fig. 1, is based on principles developed by BBC engineers to enable automatic gain control to be applied to high-slope pentodes capable of much greater gain than the variable-mu types generally used for a.g.c. The advantage of this high gain is that the use of a second r.f. stage -a notoriously unstable arrangement-is avoided.
A.G.C. by grid injection is not practicable with high-slope pentodes but it is possible to control the mutual conductance-and hence the gain-by applying sufficient negative bias to the suppressor. This inhibits the arrival of electrons at the anode and compels many (all at full cut-off) to alight upon the screen, where they are unable to influence the external circuit. As a rather high control potential is required, some d.c. amplification is necessary and this, fortunatcly. can be provided by the detector.
At this stage, as audio
quality is one of the

vith amplified a.g.c.
objects, the question of detector distortion must be put into correct perspective. So far as background silence and audio bandwidth are concerned v.h.f. is supreme, but the harmonic distortion of the usual discriminator, at about $3 \%$, is rather higher than that of some a.m. detectors. (I know of only one somewhat rare, though simple, f.m. discriminator that can maintain a truly distortionless performance.)

The figure is horrifying but complaints are few even among high fidelity enthusiasts and we can bear this in mind when balancing the interests of sensitivity and quality.

The distortionless a.m. detectors have less than unity gain and require some r.f. amplification even for local station reception. Additional r.f. amplification to improve the sensitivity is almost certain to bring distortion and instability to cancel all the advantages of such a detector. Better results are likely with a detector that can itself contribute enough gain to work well with the stable single r.f. stage.

Although an anode-bend detector may distort on occasional heavy modulation, its performance on average signal is good. With a high-slope pentode a gain of over 100 is possible and the stage is particularly suitable as a source of amplified a.g.c. potential. To avoid loss of d.c. gain through negative feedback in the cathode resistor R4 it must be kept small (a bypass capacitor would not prevent this feedback from reducing the gain of the d.c. component of the signal).

Nevertheless the resistor must still yield the correct bias and this is achieved by making it carry additional current to increase the voltage drop, the current being that of V1, which has the earthy end of its cathode network common with that of V2.

For a similar reason the screen of V 2 derives a stable potential from the cathode of V1, which is raised rather high above earth to suit the a.g.c. conditions. The d.c. control potential from the anode of $V 2$ is filtered by the network C5, R6, C6
to eliminate the r.f. content and passed via the a.f. filter C7, R 7 to the suppressor of the r.f. valve.

It is, of course, a positive potential varying with the signal strength but must be negative with respect to the cathode of V1. It is for this reason that the $12 \mathrm{k} \Omega$ resistor R 3 is included to raise the cathode. The normal potential difference for grid bias is picked off by returning the grid resistor R1 to the junction of R2, R3. The grid capacitor C2 is essential to block the d.c. path to earth through the tuning coil.

## Choice of Valves

Choice of r.f. valve is governed by its suppressor grid base, for this must be short enough to lie within the range of available control voltage. My admittedly brief experiments seemed to indicate that the EF50 recommended by the originators of this a.g.c. system is more suitable than some newer types, but there is scope for the interested amateur to take the investigation further. As the veteran EF50 is readily obtainable as a surplus item at less than half a crown it is included in the present unit.

Although this valve seems large by today's standards, and other components used were by no means the smallest made, the tuner was built on a "bread board" measuring 6 in . x $3 \frac{1}{4} \mathrm{in}$. without crowding. The detector is an EF80 but type EF50 can be used for this stage, too.

Ordinary aerial and r.f. coils are suitable for a circuit such as this, in which knife-edge selectivity is undesirable, and it is not necessary to limit the constructor to one specific make. They should be canned because of the high gain of the circuit. The base connections of coils may not be the same as indicated in the point-to-point wiring diagram (Fig. 2) but this will necessitate only'slight variation of the wiring layout.
The potentiometer in the detector cathode circuit is included to enable the a.g.c. operating conditions to be adjusted. Once set it can be forgotten, but experimenters who wish to exploit the full gain of the unit could mount this component on the front for use as a sensitivity control.

It is usual to power a small unit *such as this from an outlet in the main amplifier, and therefore no supply arrangements are included here. It is unlikely that the tuner will be granted the efficient aerial so often recommended for a good signal-tonoise ratio and it must be admitted that a really large one would bring selectivity troubles with the simple tuned circuits employed. In the Midlands the customary bit of wire-about $2 y d$ dangling behind the tuner-brings in several Continental transmissions at good strength and quality.

Conventional chassis construction could be adopted but a " bread board" layout can be made compact, reliable and requires no metalwork.

Using available components rather than the smallest possible, the prototype was assembled on a piece of unmetallised laminated plastic 6in. $x$ $3 \frac{3}{8}$ in. $x \frac{1}{16}$ in. thick.

The wiring was arranged as shown in the point-to-point diagram, Fig. 2. The potentiometer VR1 is under the panel and occupies more depth than is generally needed for bread board wiring, but this depth permits C3, C7 and C8 to be housed below to avoid increasing panel area.

The two-gang capacitor. canned coils, standard valveholders, aerial soc:iet and audio output socket are all mounted on top and solder tags are fitted under their fixing screws to enable earth continuity wires to link them. Small trimmers should be fitted if there are none already present in the tuning capacitor.

## Wiring-up

Begin wiring by soldering bare wire from tag to tag on the top side to form the earth links, remembering to include the frame of the tuning gang and the coil cans, which are not earthed automatically as in the case of metal chassis construction. Drill small holes at convenient points to enable short earth connections to be passed through to the underside and soldered to the appropriate tags on the valveholders, coils and potentiometer as shown in the diagram.

Next connect the heater wiring, which has been omitted from the diagram for clarity. As this will be taken from the main amplifier it will probably
be a centre-tapped supply requiring an insulated twisted pair. For an unbalanced supply (one side earthed) run a single insulated wire to one heater tag on each valveholder and earth the other tag. No socket is fitted for power supply connections: instead the heater, h.t. and earth wires are led out as flexible leads ending at a plug to suit the outlet on the main amplifier.

Most of the circuit wiring is in 20 s.w.g. bare tinned copper wire "stitched" to the panel by passing it through small holes. Anchorage, where needed, is provided by passing wire twice through close pairs of holes and, bonding solid by soldering.

At first glance the diagram may seem to show breaks in the coil-to-grid connections, but use is made of the tags provided on both sides of the fixed vanes in most tuning gangs so that the path to the grids is continuous across the vanes. Insulated wire is shown at points where the diagram might suggest contact in unwanted places but in practice these could be air-spaced and bare wire used.

The prototype has an epicyclic tuning drive, this being simple and reliable and free from the troubles so likely to occur with cord and pulley arrangements. As the interest is in only those transmissions that are received well it was decided that a ready-made dial printed with all stations from here to Tokyo would be of little benefit. The home-made one used instead has a limited number of stations marked to indicate three standards of reception quality.

Although it was intended to add the station


Fig. 2: Underside view of the "bread-board" wiring.

| COMPONENT LIST |  |  |  |
| :---: | :---: | :---: | :---: |
| Resistors: |  |  |  |
| RI | $470 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W}$ | R5 | 270k $\Omega \frac{1}{2} \mathrm{~W}$ |
| R2 | $150 \Omega \frac{1}{4} \mathrm{~W}$ | R6 | $100 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W}$ |
| R3 | $12 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W}$ | R7 | $1 \mathrm{M} \Omega \frac{1}{4} \mathrm{~W}$ |
| R4 | 470 $\frac{1}{2} \mathrm{~W}$ | VRI | $\mathrm{l} \Omega \Omega$ pot. |
| Capacitors: |  |  |  |
|  | C4 2-ga | 2-gang 500 pF |  |
| C2, C5, C6 100pF |  |  |  |
| C3 |  | $16 \mu \mathrm{~F}$ electrolytic 150 V |  |
| C7 | $0.1 \mu \mathrm{~F}$ |  |  |
| Valves: |  |  |  |
| VI | EF50 | V2 | EF80 |

names the users so quickly memorised the transmissions associated with the simple dash marks that even that refinement has been omitted! The usual dial lighting arrangements can be included if required.

Alignment of the tuned circuits follows the usual procedure for a t.r.f. tuner and can be done "by ear". The split-vane type of gang permits a useful correction for tracking error that might be evident at the ends of the band, but if overdone there is a risk of contact between fixed and moving vanes. In the conditions that have to be accepted
as normal for the medium-wave band nowadays, only a few of the receivable transmissions are acceptable as entertainment for the music lover.

When these have been logged the split vanes can be used to favour them. The potentiometer has "dead" and "live" positions at each end of the track. Set it at the live end and then adjust it until a weak signal just begins to fall in strength. A.G.C. will then be operative.

If you are more interested in quality than in logging the maximum number of stations it is worthwhile to reduce the sensitivity still further with this control to the point where the best signals are still comfortably received. This avoids overloading the detector and also keeps the unit free of the slight positive feedback that can degrade quality at the highest sensitivity levels even though still clear of oscillation.

The quality obtainable is about the best that is likely from distant medium-wave stations but the unit cannot restore audio frequencies missing from the transmission nor adjust for the overmodulation that is evident in some broadcasts. Whistles can sometimes be handled by the filters and tone-control circuits of the high-fidelity equipment. However, it is generally possible to log a number of stations that can provide an agrecable standard of entertainment.


## An unbeatable machine based on the game of Nim




By J. R. Stewart

NIM is a game for two players. Starting with a number of matches, counters or other objects arranged in piles or rows, each player alternately draws as many matches as he chooses from one row only. Two variations of the game are possible; in one, the player who picks the last match wins the game, in, the other, he loses. There is no limit to the number of counters in any one row, nor to the number of rows with which to begin the game, although recent TV programmes have made popular the game of "last match loser" with four piles containing 7,5,3 and 1 matches.

## The Theory of Nim

The theory of this game has been completely worked out, and an elementary knowledge of it helps to understand the construction of the machine. To begin with, the operation of "last match wins" variation will be explained.

Briefly, any number can be written as the sum of some of the powers of 2 , together with the number 1 , i.e. using the numbers in the sequence $1,2,4,8,16$, etc. For example, $9=8+1 ; 30=$ $16+8+4+2$. In representing any number by such a sum, none of the sequence numbers occurs more than once. This can be checked by noting that the sum of all the sequence numbers up to any point is always 1 less than the next number in the sequence; for example, $1+2+4=7$, and the next sequence number is 8 .
Calling the individuals Player and Opponent, it can be shown that in playing Nim, if the Player produces what we call a safe combination, the Opponent's move always upsets this arrangement. allowing the player to reset a safe combination when he draws in his turn. By safe combination we mean a set of numbers the numbers of counters in each.row) which will lead to a win lor the Player. The secret of the game lies in
knowing the safe combinations, and how to produce them. For "last match wins" variation, th safe combinations are found as follows:
(1) Represent the number of counters in eacl row as the sum of the sequence number given above;
(2) Count the number of $1 \mathrm{~s}, 2 \mathrm{~s}, 4 \mathrm{~s}, 8 \mathrm{~s}$, etc. whicl appear in the arrangement;
(3) A safe combination is one in which th numbers of each of the $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. is EVEN and the Player who sets up such a combina tion will win.
Take the familiar example of $7,5,3$ and 1.

| $7=4+2+1$ |
| :--- |
| $5=4+1$ |
| $3=$ |
| $1=$ |$\quad 2+1$

Reading vertically we see that the numbers o $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. are all even, hence this is a saf combination for the Player who set it up. Whe the Opponent draws, because he draws from on pile only, he must make some of the numbers of $1:$ 2 s , etc. odd, since he can remove only one 1,2 or 4. This makes the combination unsafe, and th Player has then only to even things up again $t$ keep a winning line. Suppose the Opponen removes all the 7 counters from the first row. Thi then leaves

$$
\begin{array}{ll}
5=4 & +1 \\
3= & 2+1 \\
1= &
\end{array}
$$

All the $1 \mathrm{~s}, 2 \mathrm{~s}$ and 4 s are now odd. To even th arrangement, the 4 must go, so that the Playe must draw from the 5 row. The other rows ar even in 1 s and odd in 2 s . Thus if he leaves onl 2 counters in the 5 row, the combination will $b$ safe. i.e.

$$
\begin{aligned}
& 2= \\
& 3= \\
& 1=
\end{aligned} \quad 2+1
$$

When the machine plays, the procedure is that it is "told" the, numbers of counters in all but one of the rows. These numbers are given to it in the form of the sequence numbers $1,2,4,8$, etc. It then counts the numbers of $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. or rather it selects the cases in which these numbers are odd. This selection is then the number of counters which the machine decides to leave in the last row. Again, an example will make things' simpler. SLppose three rows contain 9, 7 and 4 counters.

$$
\begin{array}{ll}
9=8 & 4+2+1 \\
7= & 4
\end{array}
$$

This is odd in 8 s and 2 s , so that the machine will elect to have $8+2=10$ in the fourth row, making its own safe combination. If the opponent draws ' 5 from the 10 row, this leaves-

| $9=8$ | $4+2+1$ |
| :--- | :--- |
| $7=$ | $4+2$ |
| $4=$ | 4 |
| $5=$ | 4 |

If the mashine is again told the first three numbers, it will again select 10 for the last row, which is now impossible, since 10 cannot be left
that the machine refuses to play.
Thus when playing with $7,5,3$ and 1 , which we have seen is a safe combination, if we tell the machine 7, 5 and 3, it will reply 1 : if we tell it. 5,3 and 1 , it "will reply 7 and so on...To sum up, the machine is unbeatable if it is allowed at the start - of the game'to select the number in one row, having been told the numbers in the others.


Fig. 2: The bosic switching circuit.

If the opponent sets out the numbers in all the rows at the start, the machine will beat him if he sets an unsafe combination. If he sets a safe combination, the mashine will refuse to draw from any pile; this could be called a drawn game.


Fig. 1: The layout of the front panel.
if there are only 5 in the row. Looking at the last three, the combination is odd in 4 s and 2 s . Thus if the machine is told the numbers 7,4 and 5 , it will select $4+2=6$ for the other row, i.e. it draws 3 from the 9 row, again making its own safe combination. The machine does not always draw from the row having the largest number of counters; if the rows contain $9,8,7$ and 4 it will reduce this to $9,8,5$ and 4.

With a four-row machine, there are four possible ways of selecting any three rows, i.e. there ate four possible sets of numbers which can be told to the machine. One of these will always give a workable answer-workable in the sense that the number selected by the machine will be less than the number already in the remaining row. This applies provided that the combination is unsafe. If the combination is already safe. no matter what selection of three from four is made, the machine will always reply with the number already in the remaining row; in this sense we can say

## Construction

Using four rows the machine requires to be told the numbers in three rows. This is done by throwing yarious switches. The reply which the machine makes is read off on small lamp bulbs. The $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. of the sequence numbers are completely independent, and depending on the number of switches one uses, the machine can be made to cope with any given number in a row.

Calling the rows $\mathrm{A}, \mathrm{B}, \mathrm{C}$ and D , and designing for up to 15 in a row, we need (Fig. 1) 12 switches, and four lamps. The number of counters in rows A, B and C, are told to the machine by throwing the appropriate switches in each column; the lamps which light up in row D tell us how many counters to leave in that row Remembering that a safe combination has an even number of $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. and that the $1 \mathrm{~s}, 2 \mathrm{~s}$, etc. are independent of each other we get the following set of rules which the switching circuit must give:
Reading horizontally along any row,
(1) None of the switches A, B or C, thrown must keep the lamp off;
(2) Any one of the switches $\mathrm{A}, \mathrm{B}$ or C , thrown must bring the lamp on;
(3) Any two of the switches A, B or C, thrown must keep the lamp off;
(4) All three switches A, B and C, thrown must bring the lamp on.
In this way we always have an even number of (lamp + switches) in the ON position.
A switching circuit which will perform these operations is shown in Fig. 2. As drawn, all the switches are in the OFF position, and the arrower'
contact moves across to the "free" contact when the switch is thrown. Each switch and lamp has been marked to correspond with the row in which it is situated. If the connections are traced out, it will be seen that conditions 2 and 4 bring the lamp on, while conditions 1 and 3 provide no conducting path for the supply. From Fig. 2, the switches in row A require a single change-over contact; those in B and C a double change-over contact.


Fig. 3: The circuit of the bottom row of switches and power supply.

Toggle switches having a double change-over can be purchased; because the writer required a machine which would play both varations, and as "last match loses" needs an extra contact on nearly every switch, Post Office type key switches were used in the unit to be described. These have the added advantage that the handle, being longer than the toggle dolly, is more easily read to determine a given number.

Once again it should be emphasised that there are no connections between the horizontal rows of switches except that of a common supply line. A 3 V dry battery can be used, together with 2.5 V bulbs. One side of all the bulbs in D will be connected together to one side of the battery, while the other battery terminal, through an ON/OFF switch if desired, goes to the movable contact point of all the A switches.

## Last Match Loses

The above is the simplest version of an unbeatable Nim player, playing last match wins. For those who want to play last match loses, or both variations, the circuit becomes more complicated. In playing last match loses the theory of the game is only slightly different. The same combinations as in last match wins are safe, until the game reaches the stage of only 1 or 0 in all the piles. Then an odd number of is is safe, while an even number is unsafe. A little reflection will show this to be true; if there are three singles matches left, the opponent who picks the first of these must also pick the last, thus losing the game.

We therefore require a circuit which is the same as in Fig. 2 for the top three rows, i.e. 8s,


Fig. 4: Details of the front panel. The four holes for the keys ore drilled in tin plate only, not in the cover.



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Fig. 5: Details for mounting the lamps.
4 s and 2 s , and which will reverse the lamp operation in the bottom row of is if, and only if, all the switches in the $8 \mathrm{~s}, 4 \mathrm{~s}$ and 2 s , are in the OFF position. Fig. 3 shows the circuit for the bottom row of switches, and the power supply. A changeover contact on a relay is used to reverse the operation of the bottom row, the relay being controlled by the pcisition of the keys in the 8 s , Is and 2 s . When the relay is off, is drawn. Fig. 3 gives the same conditions as Fig. 2. When the elay closes. the right-hand coriact of switch A comes into speration, and we then get
(1) None of $\mathrm{A}, \mathrm{B}$ or C , hrown, lamp lights;
(2) Any one of $\mathrm{A} . \mathrm{B}$ or C , hrown, lamp off;
(3) Any two of A, B or C, hrown, lamp lights;
(4) All three of A, B and C, hrown, lamp off.
These are the safe combinations or last match loses when educed to only one match in a ow.
The relay is controlled by switch ; and by nine contacts in series, me on each of the $8 \mathrm{~s}, 4 \mathrm{~s}$ and 2 s witches. These contacts are losed when the switches are off; when any one of these switches

'g. 6: Mounting and wiring of power supply and lamps.
is thrown, the relay is kept off. Switch S is brought out to the front panel; if it is open the relay remains off and the machine plays last match wins, if closed the machine will play last match loses.
Convenience and price dictated the design of the power supply; batteries were discarded because of the trouble of replacement, and also because a rather higher voltage is usually needed for a relay than for the lamp bulbs. A search through the catalogues brought up a 6 V operating relay with a change-over contact, and it was then decided to use a valve filament transformer together with 6 V dynamo bulbs, each taking $0 \cdot 3 \mathrm{~A}$. If it is felt that these may be overrun on 6.3 V , and there is a tendency sometimes to be over-generous with filament voltages, particularly as the transformer is not working at full load-a $6.3 \mathrm{~V}, 2 \mathrm{~A}$ rating transformer is used-a length of


Fig. 7: The wiring of keys in $8 s, 4 s$ and 2 s rows. The upper half of each key carries the relay control switch; the lower half controls the lamp in the $D$ pile.


Fig. 8: Wiring the keys in /s row.
resistance wire can be incorporated in each lamp lead. The direct supply to the relay is obtained from a battery charger type half-wave rectifier, rating $6 \mathrm{~V}, 1 \mathrm{~A}$ charging a $50 \mu \mathrm{~F}, 12 \mathrm{~V}$ electrolytic. This gives an off load voltage of 8 V , dropping to just under 6 V when the relay comes in . No mains switch has been used, the unit being on when it has been plugged in. This is permissible since it is unlikely that it will be left plugged in when' not in use.

## Housing the Instrument

The writer already had a wooden box with hinged lid, measuring $15 \frac{1}{4} \mathrm{in}$. $x 9 \frac{1}{4} \mathrm{in}$. $x 4 \mathrm{in}$. deep. and this was used to house the unit. Any other similar size will serve, the only critical dimension being the 4 in . depth, as the keyswitches require this amount. Lid and base were removed, and a front panel $15 \frac{1}{\mathrm{i}} \mathrm{in}$. $x 9 \mathrm{in}$. of tin plate was cut and drilled as shown.
Two small wooden supports, 1 in . $\times 1 \frac{1}{8} \mathrm{in}$. $\times \frac{1}{2} \mathrm{in}$. were glued one on each side of the box opposite

[^5]
# THE HOME PRODUCTION OF <br> Prinled Ciccuil Boarids DESCRIBED BY H. W. RUTT 

PRODUCING printed circuit boards in a home workshop is not as difficult as it may sound. This type of circuitry is very convenient for miniature units, and it practically eliminates wiring errors.

The first step, naturally enough, is to design the layout of the conductors on the circuit board. Until some practice is gained in designing these boards it will be found rather difficult. The only tips one can give are not to have the conductors spaced by less than $\frac{1}{10} \mathrm{in}$., and not to have the conductors narrower than to in. It should also be remembered that the circuit will have to be painted out by hand, and so should not be impossibly complicated.

## Drawing the Circult

When the design has been finalized it should bedrawn out on a piece of tough paper of the correct size. The copper-clad laminated plastic board is then cut to size, using a fine-bladed hacksaw, and cutting with the copper layer uppermost. When the board has been cut, its edges should be smoothed down with a fine-toothed file to remove the rough edge of the foil. The design is now transferred to the coppered side of the board by means of carbon paper.

The areas where the copper must remain are now painted with a black, cellulose paint, of the type sold by garages, and this is allowed plenty of time to dry. It is wise to check the painted areas for tiny holes in the paint caused by air bubbles, and to check the layout of the conductors against the design.

## Etching Process

The next step in producing the hoard is to etch away the exposed copper. The etching hath consists of a solution of ferric chloride with $1 \%$ by rolume of dilute hydrochloric acid added. A $10 \%$ solution of ferric chloride may sometimes be
obtained from the chemist without waiting. How ever this $10 \%$ solution gives a rather slow etch and if one is prepared to wait a few days th chemist will usually be able to obtain, or make uf the correct etching solution, which is a $30 \%$ solution of ferric chloride in water, with $1 \%$ o dilute hydrochloric acid (i.e. 30 gr of ferric chlorid made up to 100 cc with distilled water and lcc o dilute hydrochloric acid). This solution will no burn skin (though it should be quickly washed of however, it stains terribly, and old clothes shoul be worn when using it.

The board, with the design painted on it, i placed in a shallow tray of the type used by photc graphers, and just covered with the etching solu tion, which may be warmed to about $40^{\circ} \mathrm{C}$; it mus not, however, be hot. The tray is now gentl rocked until all the exposed copper has dissolver When this has happened the board should be we washed. and then allowed to dry.

When the board is dry the paint used to proter the conductors has to be removed. The hest sut stance for this job is trichloroethylene, which ma be ordered from the chemist. It costs about 6 s . fo a $\frac{1}{2}$ litre jar. If one does use this substance, car should be taken not to inhale much of its vapou as it has harmful effects. It is possible that othe substances such as turps or petrol could be used t soften the paint, but if something like this is tries it should be checked (on an old piece of boarc that it does not damage the bakelite backing, $c$ cause the copper to peel off.

Whatever substance is used, the board should $b$ allowed to soak in it until the, protective paint soft; and then it should be gently brushed off.

## Drilling the Board

The final operation in preparing the board is $t$ drill it. Each point where a hole is needed marked, but not centre punched; centre punchin can easily cause the backing to crack. A No. 5 : slow spiral drill is used to make the holes for wirt which have to be soldered to the copper. Th board must be drilled from the coppered side, 1 prevent the copper peeling off the board, and slow spiral drill is essential for this operation.

The board itself is now finished, but sever: points should be observed when wiring it. Firstl: all connections to the board should be mac quickly, or the heat may cause the copper col ductor to peel off the bakelite backing. Second one should be very careful to solder in mult contact parts (such as IFT's) the right way roun the first time-they can be very difficult indeed 1 remove if you do not get it correct. Thirdly a resistors, capacitors, etc., should be right u against the bakelite backing, for if they are ni any pressure on the component will tend 10 pe the copper off the board. When the board is con pletely wired up and tested the soldered side of it board niay be varnished to protect it from co rosion, but this tends to make servicing rather mo. difficult.

Any constructors wishing to make their ou printed circuit boards should have little difficul obtaining the chemicals or the copper-cla laminate board, and it is hoped that this article wi help to bring printed circuits out of the "oh, that much too difficult" class of construction.

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##  Variable Mains Transformer

$12,15,20$ and 24 V . A fraction of the secondary voltage is added to, or subtracted from, the mains voltage. This gives a range of 60 V . However the primary has tappings for 240,220 and 200 V . These enable us to use auto-transformer action and to obtain a wider range. When the tappings on the primary and on the secondary are fully used, the output can be varied from 155 to $308 \vee$, with the mains voltage at 225 V .

The wiring is such that, if necessary, the secondary can be quickly disconnected from the primary; it is then available for supplying voltages less than 30 V . A voltmeter is incorporated in the instrument, with a switch which provides a reading of the incoming mains, the outgoing mains, or the low voltage secondary.

In Fig. 1, P1 and P2 represent voltage-adjusting panels of the type frequently associated with mains transformers. S1 and S2 are single-pole, 3-way switches. They are not ganged.

## Principle of Operation

Let us first assume that the incoming mains voltage is 220 V , and that the studs in the voltageadjusting panels, P1 and P2, are both set to 220 V . If switches S1 and S2 are both set to position $\mathbf{A}$, the outgoing mains voltage is the same as the incoming. For simplicity, only one tapping is shown on the secondary. If $\mathbf{S} 1$ is set to $\mathbf{A}$ and $\mathbf{S} 2$ to $\mathbf{B}$, the voltage $A B$ may be added to the primary voltage. Then, if S1 is set to B, and S2 to A, the voltage $A B$ is subtracted from the incoming mains

WHEN it is required to apply a fraction of the mains voltage to a given load, it is possible to use a resistive potential divider, but of ten the resistors will be bulky, and there may be problems of heat dissipation. If the supply is a.c., a variable voltage mains transformer is frequently more convenient; also, in addition to providing a ratio less than unity, the transformer can step up the voltage, which is beyond the scope of the resistive potential divider.

## Commercial Equipment

Variable mains transformers are commercially available, and are known by various trade names. They work on the principle of an autotransformer: a manual control determines the position of a tapping on a toroidal winding. It enables adjustments to be made in such small increments that the transformers are described as being infinitely variable. They can also be described as rather expensive.

The device discussed in this article uses a comparatively cheap battery-charging transformer; and it is not infinitely variable. The increments of voltage are however small enough for most purposes. A suitable transformer is the Douglas MT3. It has a 30 V secondary, tapped at


Fig. I: The circuit of the tronsformer.
voltage. Thus, adjusting the secondary switches S1 and $S 2$ gives a range from $220+30 \mathrm{~V}$ to $220-30 \mathrm{~V}$, i.e. from 250 to 190 V .

Suppose we want a voltage higher than 250 V . We shall continue to assume that the incoming mains voltage remains at 220 V .

We screw the stud in P1 in the 200 V position and that in P2 in the 240 V position. The full secondary voltage is now more than 30 V , and it can be added, by suitable operation of $S 1$ and $S 2$, to a voltage which is now greater than 220 V , because of the auto-transformer action in the primary. It is on
this basis that the outgoing mains voltage can be stepped up to a value of the order of 300 V .

If we want a voltage which is less than 190 V , we screw the studs in P1 in the 240 V position, and that in P2 in the 200 V position. S1 and S2 are adjusted so that the secondary voltage subtracts from the primary voltage.

In the model made by the writer, sufficient variation was obtained for his purpose by adjusting switches S1 and S2 on the front of the instrument. The voltage-adjusting panels P1 and P2 were mounted at the back. If greater variation is frequently required, it may be more convenient to have P1 and P2 in the form of two single-pole, 3-way switches on the front.

The secondary voltage, as selected by switches S1 and $S 2$, is taken to a 2-pin, 2A socket, K2. A 2-pin, 5 A plug Pg 1 , with the pins connected by a short length of wire, connects the primary to the secondary. It fits into the socket K1. The 5A plug has a small paxolin panel associated with it so that when the plug is inserted in K1, the $2-\mathrm{pin}, 2 \mathrm{~A}$, socket K2 is masked by the panel. It is thus impossible to insert a plug into the socket for the secondary, when the secondary is connected to the mains.

## The Voltmeter

The writer used a meter scaled from 0 to 300, which had a fúll scale deflection of 5 mA . A 5 mA rectifier was used with it. The voltmeter had two ranges: $0-30 \mathrm{~V}$, for measuring the secondary voltage; and $0-300 \mathrm{~V}$ for the incoming and outgoing mains voltage. The series resistors to be used with the meter depend on the full-scale deflection of the meter, and on the range. The following will show the method of calculating their values. The example relates to a 5 mA meter used on the 300 V range.


A view of the author's unit.

When a current with a mean value of 5 mA flows through the meter giving full scale deflection, the maximum current in each half cycle is $5 \times \pi / 2 \mathrm{~mA}$.

If the r.m.s. value of the voltage is 300 V , the maximum voltage in each half cycle is $300 \times \sqrt{ } 2$. Thus the resistance in the circuit must be

$$
\begin{aligned}
\frac{300 \sqrt{ } 2}{5 / 1,000 \times \pi / 2} & =\frac{300,000}{5} \times \frac{1}{1 \cdot 11} \\
& =54,000 \Omega
\end{aligned}
$$

For the 30 V range the resistance of the circuit must be $5,400 \Omega$. The resistance of the meter itself (and the rectifier) is negligible in comparison with these values. The voltmeter circuit is shown in Fig. 2.

The terminals at the top of the instrument are provided so that the 30 V range of the voltmeter can be used to measure external voltages. Alternatively, the external voltage can be applied by inserting a plug into a socket situated between two of the terminals. (This socket is not the socket K2, which is masked by the paxolin panel.) The voltmeter range switch, S4 in Fig. 2, is next to the meter. Switches S1 and S2 have pointer-type knobs. Of the remaining two


Fig. 2i The voltmeter circuit.
switches, one is the on/off switch S3; the other is not used in the final form of the instrument. The model is larger than is necessary for the facilities mentioned in this article; it includes a $12 \mathrm{~V}, 2 \mathrm{~A}$ rectifier, and a 3 -pin, 2A socket, for the rectified voltage.
To measure the secondary voltage selected by switches S1 and S2, the socket K2 must be connected to two of the terminals at the top of the instrument.

## Resistance and Capacitance Measurement

The provision of a link Pg1 to connect the primary to the secondary enables the instrument to be used for measuring resistance and capacitance. With the link Pgl in position, the outgoing mains voltage is adjusted to the same value as the incoming mains voltage, $\mathrm{V}_{1}$. The resistor or capacitor to be tested is then connected in place of Pg 1 . The voltage $V_{2}$, read when the voltmeter switch is thrown to the position for the outgoing mains, is then observed to be less than $V_{1}$. The ratio $V_{2} / V_{2}$

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and the value of the voltmeter resistor $\mathrm{R}_{1}$, enable the value of the resistor or capacitor under test to be calculated.

If $R x$ is the value of the resistor which replaces Pg 1 it can be shown that

$$
\begin{equation*}
R x=R^{1}\left(\frac{V_{1}}{V_{2}}-1\right)=R^{1}\left(\frac{V_{1}-V_{2}}{V_{2}}\right) \ldots \tag{1}
\end{equation*}
$$

If Xc is the reactance of the capacitor which replaces Pgl , it can be shown that

$$
\begin{equation*}
X c^{2}=R_{1}^{2}\left[\left(\frac{V_{1}}{V_{2}}\right)^{2}-1\right] \ldots \tag{2}
\end{equation*}
$$

If the frequency is $50 \mathrm{c} / \mathrm{s}$, we have

$$
\begin{equation*}
C=\frac{10^{6}}{314 R_{1}} \frac{V_{2}}{\left(V_{1}^{2}-V_{2}^{2}\right)^{\prime}} \mu F \ldots \tag{3}
\end{equation*}
$$

These expressions assume that when resistance is being measured, the component under test has no reactance, and when capacitance is being measured, the component has no resistance.

Alternatively, with the mains voltage at a value which is usual for a given site, various resistors of known value can be connected in place of the link Pg1. The corresponding values of $V_{2}$ are noted. A graph can be drawn relating $R x$ and $V_{2}$. When an unknown resistor is connected, $V_{2}$ is noted, and the corresponding value of $R x$ is read from the graph.

A similar graph can be constructed for capacitance.

## Constant Output Voltage

The action of the variable transformer has been described in terms of deriving a variable voltage from a constant mains voltage. It may be worthwhile mentioning that it can also be used to give a constant output voltage when the input voltage varies. The equipment was orginally devised for this purpose. Soon after it was completed, the local supply company improved the stability of its mains voltage.


In fact, if any readers are troubled with a fluctuating mains voltage, it may be worth their while to make up a unit of this kind. When they have finished it, they are pretty sure to find that their supply has teen improved. If however, the supply company reacts slowly, when the unit has been completed, and if some readers actually need to use the unit, they may like to know that operation is very simple: you switch on, turn the voltmeter switch to the position for the outgoing mains, and adjust S1 and S2 until you have the desired voltage. You then switch on the load.

Incidentally, with a secondary tapped at 12,15 , 20,24 and 30 V . the following increments are available: $3,4,5,6,8,9,10,12,15,18,24$ and 30 V .

These values can be obtained without using the voltage-adjusting panels. If these panels are used, fractions and multiplies of these values are a vailable.

## AN ELECTRIC NIM MACHINE

## -continued from page 529

the line of lamp holes, to support a wooden plywood battens $8 \mathrm{tin} . x \mathrm{lin}$. on which were screwed the four M.E.S. lampholders. This brings the bulbs flush with the front panel holes. One of each of the lampholder fixing screws was used to anchor a lin. wide strip of tin plate which also extends to the front panel; this serves to shield the hole from the light of the bulbs adjacent to it. A similar batten of tin. thick wood, screwed to two supports $3 \frac{1}{4}$ in. $x$ lin. $x \frac{1}{2}$ in. supports the transformer, rectifier and condenser. The relay is screwed direct to the side of the box, counter-sinking the fixing bolts from the outside.

All wiring is done from the rear with the front panel screwed into position, the relay being mounted with its contacts forward, so that the soldered ends of these contacts are readily
accessible. The wiring diagrams assume that the keyswitches have a central OFF position and a down ON position, i.e. all the key contacts shown are above the centre line of the keys.

The keys drawn have four change-over contacts. arranged in two side-by-side banks. When off, the lower two of each trio of contacts are connected. and when switched on, the centre contact is lifted to connect with the top one. In the 1s keys, only two change-overs are needed on each switch, so that side-by-side contacts can be wired together. rather than leave one set unused. This serves as a precaution against a dirty contact on one set. On the $8 \mathrm{~s}, 4 \mathrm{~s}$ and 2 s keys, another contact is needed to control the relay, and these connections have been drawn on the upper trio of contacts, again wiring the two side-by-side in parallel.

The unit was completed by cutting a cover of polished sheet (Formica was used by the writer. but paxolin will do equally well) to the same dimensions as the front panel, with lamp holes and slots only for the key handles.


## Mains/Battery Receiver

FiROM Ekco Radio and Television Ltd. comes a new transistor receiver, designed to operate either from the mains or from its own internal battery. The receiver has been named "Varsity" (model number MBT425) and it sells for 19 guineas.
The eight-transistor circuit covers medium and long waves and incorporates a transformerless output stage. The makers of the "Varsity" are Ekco Radio and Television Lid., Southend-on-Sea, Essex.


This 8-transistor portable is made by Ekco Radio and Television Ltd.

## Dual-function Receiver

THE second transistor receiver made by the Ever Ready Co. (Great Britain) Ltd. to have the dual function of portable and car radio has just been announced. Called the "Sky Tourer", it has been designed to fit into a special container which is permanently installed in the vehicle and which, with the receiver inserted, automatically connects it to the car's power supply, external aerial and separate 6 in. $x 4 i n$. loudspeaker.

When removed from the container the "Sky Tourer" operates as a normal transistor portable, using its own internal battery, aerial and speaker.

The set tunes over the long and medium wave bands and the six-transistor circuit provides 540 mW output when used as a portable and 1 W when used in the car. The price of this new receiver complete is $£ 221 \mathrm{~s}$. and it is manufactured by the Ever Ready Co. (Great Britain) Ltd., Hercules Place, Holloway, London, N.7.


The "Sky Tourer" is a new dual-purpose receiver from Ever Ready.

## Portable Radiogram

O
NE of the most interesting pieces of radio merchandise to be handled recently by Denham and Morley Ltd. (U.K. distributors of imported electrical equipment) is a portable record player/transistor receiver called the "SwingAlong ".

The receiver section covers both medium and long waves and the record player will take 45 r.p.m. records. The circuit employs six transistors and provides a maximum output of 750 mW .

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THE facility of "Trace Expansion" was mentioned in the preceding article (Part 7) of this series. While we pay a lot of attention, quite properly, to the sensitivity and bandwidth of the $Y$ amplifier, and consider the provision of the $X$ amplifier as a simple means of trace expansion, it is as well to remember that for several useful applications the $X$ amplifier comes into play.

Typical of this test technique is the display of Lissajous figures, and wobbulator waveforms.
When two sinusoidal a.c. waveforms are applied to the $X$ and $Y$ inputs of the oscilloscope simultaneously, the pattern displayed will depend

Typical Lissajous figures, with the $Y$ frequency twice the $X$ frequency (a), half the $X$ frequency (b) and a ratio of $3: 2$ (c) are shown in Fig. 11.

If the mains frequency is used as a "yard-rule" to the $X$ input, via a suitable attenuator for matching the two signals to be compared, and the frequency under test, say, the $300 \mathrm{c} / \mathrm{s}$ output from an audio oscillator, a trace of six whole loops should be displayed when the two frequencies are in step, ratio 6:1, as shown in Fig. 11 (d). As can be seen from the display, when the number of loops touching an imaginary horizontal line is divided by the number of loops touching an imaginary vertical line, we have
 the ratio of Y : X.

## Involved Waveforms

Interpretation of these figures is simple when the waveforms are exactly sinusoidal and the ratio a whole number, as in the cases stated above. Considerable experience is required to determine how many loops are displayed, let alone touching the imaginary axes. when the waveforms are distorted, the phase angle varying, and the ratio an odd figure. Take for example the drawings of actual oscillograms in Fig. 12. Drawings are used here as the difficulty of photographing the more erratic of these figures makes the outline-indistinct. and for the purpose of our argument clarity is essential.

As we shall see in a moment (Phase Measurement), when the waveform $X$ is equal to $Y$ in frequency and the phase angle is $90^{\circ}$. and voltages equal, a circle is displayed. If the phase angle changes, the circle becomes an
on a number of factors. Principal among these are: relative amplitudes, frequencies and phase of the two waveforms. If the input to the $X$ timebase is a known frequency, and that to the $Y$ amplifier a variable frequency, then the display will form a number of loops, according to the ratio between them. For normal work, ratios of about $10: 1$ are the maximum practical, and there must be a definite, stable relationship for a still trace.

If the waveforms are not exact multiples of each other the patterns of loops will rotate on the screen. This may be very pretty but does not help us much, although it is possible to note the speed of rotation of the individual loops and calculate the frequency difference.
ellipse, inclined according to the phase angle.
Fig. 12 (b) and (c) shows the ellipse with sinusoidal inputs and again with a distorted wave. form. such as given by the clipping of the $X$ waveform shown in (d). Note that the first ellipse can be considered as one loop touching the horizontal and one the vertical line and thus $\mathrm{fy}=\mathrm{fx}$. But the severely distorted display also resolves to $\mathrm{fy}=\mathrm{fx}$, if a close look is given to the actual loop formation.

Going a little further, consider Fig 12(e) where the $Y$ waveform is three times that of the $X$ waveform (frequency). The phase angle has distorted the trace, but there are still three contact points on the horizontal line and one on the vertical, as indicated.


Fig. 12: Frequency comparison: (a) shows the basic connections; (b) ellipse formed when $f y=f x$ - phase angle $90^{\circ}$; (c) distorted trace of (b) when input is nonsinusoidal; (d) X-waveform producing (c); (e) $f y=3 f x$, waveform distorted; and (f) fy: $f x=4: 5$; note superimposed loops.

Even more complicated is Fig. 12(f), which indicates a ratio of $f y: f x=4: 5$, and where the superimposition of the loops can only be sorted out by careful tracing.

The point of the X amplifier can now be seen. If it is required to interpret a complicated display. and, moreover, one that may be moving all the while, it is desirable that the trace be expanded or contracted to readable limits.

## The Circular Timebase

It is evident, from some of the traces of Fig. 12, that a large ratio of the two frequencies being compared would make interpretation extremely difficult. A method of comparing frequencies with a wide difference is to apply a circular timebase. This consists of a phase-splitting network, as in Fig. 13(a), producing a phase difference of $90^{\circ}$ between points A and B .

To obtain this, R must equal the reactance of $C$, and so the frequency of the $X$ signal must be
known. If the mains frequency is used, and $C$, for example is $0.1 \mu \mathrm{~F}$, then $\mathrm{XC}=\frac{10^{6}}{2 \pi \mathrm{fC}} \Omega$ which is $\frac{10^{7}}{314}=32,000 \Omega$, which must be the exact value of $R$.
The test frequency is applied between the deflector plate system and the final anode, thereby effectively varying the deflection sensitivity, so the circular timebase is modified into a succession of ripples. Each inward and outward excursion of the spot represents one multiple of fx , so that the trace shown in Fig. 13(b), having six such ripples, denotes an $f z$ of $6 \times 50=300 \mathrm{c} / \mathrm{s}$. If a double trace is obtained the frequency ratio is $n / 2$, where $n$ is the number of loops. Thus, Fig. 13(c) has nine loops and a double trace and the ratio is therefore $9 / 2$, or $225 \mathrm{c} / \mathrm{s}$, where a $50 \mathrm{c} / \mathrm{s} \mathrm{fx}$ is used. Where a triple trace is obtained, the ratio. is $n / 3$, and so on.

Other methods of applying networks to produce an elliptical trace for easier counting of the loops. and of modulating the tube differently to produce a dotted trace, are employed for frequency comparison.

## Phase Measurement

Of more practical use to the amateur and serviceman is the hook-up used to measure phase shift. This can be applied to an audio or r.f. amplifier, an LC filter or similar network, and proves handy in checking phase shift of an amplifier, particularly of negative feedback loops. Obviously, some care is needed, both in the methods of connection and in the observance of frequency limits, as the oscilloscope amplifiers themselves can introduce some phase shift.


Fig. 13: Phase-splitting network for circulor timebase;
(b) $f z=6 f x$; (c) $f z=9 / 2 f x$ - double trace.


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In the set-up shown in Fig. 14(a), the signal applied to the amplifier input is also fed to the $Y$ input of the scope, while the signal from the output of the amplifier goes to the X input, with a correct load resistor R to maintain matching. and a variable resistor X Def.

The procedure is to apply sufficient signal to drive the amplifier fully, but not to overload it, adjusting the $Y$ amplifier gain of the scope to give about two-thirds deflection. If the $Y$ input is now disconnected and the deflection potentiometer adjusted (timebase switched off), a similar horizontal deflection can be obtained. (If the oscilloscope has the internal $X$ amplifier we discussed formerly, so much the better; the potentiometer can be dispensed with).
When the $Y$ input is reconnected, an ellipse will be displayed, its size, shape and angle depending on the phase shift between input and output of the amplifier. A sample of the ellipses that are displayed for various principal angles are shown in Fig. 14(b). Note that in each case the shift controls are adjusted so that the ellipse sits centrally about the crossover line on the graticule.

Now, if a measurement is taken from the point where the upper limit of deflection is projected across to the vertical axis, to the horizontal axis, calling this $B$; then another measurement from the point where the ellipse cuts the vertical axis, calling it A , we can calculate the phase angle from the formula Sin $\theta=\mathrm{A} / \mathrm{B}$.

It will be noted that only when the phase angle is zero (or $360^{\circ}$, which is the same thing in this case), or $180^{\circ}$, is there a straight, diagonal line. And when the phase angle is $90^{\circ}$ the trace displayed is a circle.

In practice, this also depends on matching the amplitudes of the signals, and will often require that the sensitivities of the oscilloscope amplifiers have to be known. Again, this is where the $X$ amplifier becomes useful. Note also that a variation in the supply voltage can make a difference, and a flattening of the trace in opposite quadrants, as opposed to the formation of an ellipse, indicates the presence of harmonics in the base supply.
"TV Alignment," which appeared in the May and June issues of Practical Television, and to Part 5 of this series.
The latter article, dealing with f.m. signal generators, with illustrations of some typical response curves, appeared in the July issue of Practical Wireless. Alignment of an a.m. receiver can be carried out in much the same way as described previously, with the difference that input settings and oscilloscope connections will be different. The Y input of the scope is connected to the detector load, via a screened lead. A convenient point for connection is often the top of the volume control.

## Wotbulator Deviation Frequency

The deviation of the wobbulator must be slightly more than the passband characteristics of the tuned circuits of the receiver. Too small a deviation will restrict the response curve, too great a deviation will make the displayed curve too narrow for effective observation.

The deviation frequency is derived from the $\mathbf{X}$ sawtooth waveform, the wobbulator connected to the aerial and earth sockets of the set (or to the


Fig. 14: Phase shift measurement: (a) typical hook-up; (b) phasedifference oscillograms.

## Use of a Graticule

Where an oscilloscope does not have a transparent graticule, it is essential that some form of visual indication of trace length is available. A graticule is easily made from a piece of transparent plastic, and one simple method of noting measurements is to mark the plastic where required with a soft chinagraph pencil, whose mark can easily be erased.

The foregoing is sufficient also to demonstrate the way a response curve is traced out. For further notes on alignment. the reader is referred to the companion articles to this series, entitled
output from the frequency changer if only the i.f. amplifiers are to be checked). The wobbulator is set to the central frequency either the nominal i.f. or the selected radio frequency, and the deviation about $20 \mathrm{c} / \mathrm{s}$, adjusted at the wobbulator. The X timebase is set for an adequate display, and the marker signal, if available, applied from a fixed r.f. oscillator-which can be a signal generator with the modulation switched off.

As previously described, many wobbulators have an inbuilt marker generator, often crystalcontrolled. But even if this is absent, and the


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## DERBY AND DISTRICT AMATEUR RADIO SOCIETY

Hon. Sec.: F. C. Ward, G2CVV, 5 Uplands Avenue, Littleover, Derby.
The programme for August began with a surplus sale held on the 7 th. This was followed on the 14 th by a meeting devoted to preparations for the Society's mobile rally which was hold four days later at Rykneld School.
The last event to be held in August was a direction finding practice run on the 28 th .
Another sale was held on September 4th, the first meeting of the month.

## FLINTSHIRE RADIO SOCIETY

Hon. Sec.: A. Antley. Fairholme, Fairfield Avenue, Rhyl, Flintshire.

At the meeting on August 26th, L. W. Barnes (GW3PCZT) gave a lecture on "Single Sideband". which followed the usual slow morse practice and talk on "Simple Hints and Kinks".
NORTHERN HEIGHTS AMATEUR RADIO SOCIETY
Hon. Sec.: A. Robinson, G3MDW, Candy Cabin, Ogden, Halifax, Yorkshire.
Members of chis Society manned anocher two demonstration stations during August: one on the 10th at the Halifax Agricultural Show and the other on the 17th at the Forest Cottage Community Centre Gala. August 14 th was ragehew night and on the 28 ch , a number of members brought along their photographic slides to provide the evening's entertainment.
MADIO CLUB OF SCOTLAND
Hon. Sec.: A. Barnes, GM3LTB, 7 Southpark Terrace, Glasgow, W.2.:
The constitution and financial affairs of this relatively young society are still rather involved, owing to the decision to operate the Glasgow section as a separate group within the RCS, and not as the "mother" body as had been the situation previously.
The Club's journal-"GM Magazine"-has undergone a change in cover design with the latest issue.
READING AMATEUR RADIO CLUB
Hon. Sec.: R. G. Nash, G3EJA, "Peacehaven", 9 Holybrook Road, Reading. Berkshire.

This Club reports increasing attendance figures for meetings. On August 25th another mobile pienis meeting was held at the Childe Beale Trust when mambers and their families enjoyed an informal afternoon outing.
At the last meeting for August, Dud Charman (G6CJ) gave a lecture and demonstration of aerials.
RODING BOYSSOCIETY: RADIO SECTION
Hon. Sec.: R. T. Marchant, is4 Essex Road, Leyton, London, E. 10 .

The Sociecy's annual field day was held successfully during August with the single transmitter operating on 160 m .
SCARBOROUGH AMATEURRADIO SOCIETY
Hon. Sec.: P. B. Briscombe, G8KU, "Roesacre", Irton, Scarborough, Yorkshire.
Local short wave enthusiasts and visitors to Scarborough are invited to the meetings of this club which are held at 8 p.m. every Thursday at Chapman's Yard, Waterhouse Land. North Street, Scarborough. At the meeting for August 8th, "v.h.f. workint" was under discussion.
An R,A.E. paper formed the topic for the meeting on August I 5th and a wesk later "Servicing" was discussed.

## STOCKPORT RADIO SOCIETY

Hon. Sec.: E. G. Houldsworth, G6NM, 52 Worsley Crescent, Stockport, Cheshire.

Any potential members are invited to go along to meetings of this Society, which are held on alternate Wednesdays at 8 p.m. All meetings are at the Blossoms Hotel, Buxton Road, Stockport; the next being on September llih.
STOKE-ON-TRENT AMATEUR RADIO SOCIETY
K. H. Parkes, G3EHM, 28 Grove Road, Heron Cross, Stoker on-Trent, Staffordshire.

Recent events have included a talk on aerials and a demonstration of a transistorised electric organ.
Decoration and renovation work is at present in progress at the club premises.
STOURBRIDGE AND DISTRICT AMATEUR RADIO SOCIETY
Hon. Sec.: R. A. G. Maclntosh, 50 Field Lane, Oldswinford, Stourbridge, Worceste-shire.

As it was not possible for the Society to meet at the usuat rendezvous during August, members assembled at a local hotel where a meeting was held
WESSEX AMATEUR RADIO GROUP
Hon. Sec.: G. K. Fowle, 138 Surrey hoad, Brankeome, Poole, Dorset.
The only meeting for August was held on the 12th, when mem=" bers heard a talk ziven by R. Weston.
G. J. Fowle's lecture of September 2nd was entitled "Building Mains and Battery Operated Communication Receivers".
WEST KENT AMATEURRADIO SOCIETY
R. Trevitt. 28 Delves Avenue, Tunbridge Wells, Kent.

On August 25 th members and cheir families were able to enjoy a picnic, arranged by the Society and held in the beautiful grounds of Sheffield Park.
R.S.G.B. Contests for September. V.H.F. National Fiald Day (September 7th to 8th); D.F. National Final (September 15th) and Low Power Field Day (September 22nd).

## COURSES OF INSTRUCTION

Below are listed a number of technical institutes which will be running radio and television courses this autumn. All courses begin in September (except where stated) and furcher details Can be obtained from the institute itself or from the address given. Allan Glens School, Montrose Street, Glasgow: radio theory, morse, G.P.O. regulations, aeriais, ete., and a general radio course. Bradford Technical College: inquiries-Mr. D. M. Pratt, 30 Lyndale Road, Eldwick, Bingley, Yorkshire: R.A.E. and morse courses.
Brentford Evening Institute: inquiries-Evening Institute Department. Education Offices, Town Hall, Chiswick, London, W.4: radio amateurs' course, morse, radio and TV servicing.
Carshalton College for Further Education: inquiries-The Registrar, Carshalton College for Further Education, Nightingale Road, Carshalton, Surrey: if enough interest is shown radio theory and morse classes will be held.
Central Evening Institute, Lea Mason Centre, Bell Barn Road, Birmingham: R.A.E. courses.
Hendon College of Technology, The Burroughs, Hendon, London. N.W.4: short courses on digital computors, eransistor circuit design (beginning Oetober), microwaves, global communications (beginning October) and high fidelity sound reproduction (beginning Ocrober), also a design course for electronic engineers (beginning October).
Holloway L.C.C. Evening Institutes: inquiries-A. W. H. Wennell, 145 Uxendon Hill, Wembley Park, Middiesex: R.A.E. and morse classes.

North End Evening Institute, Portsmouth: inquiries-The Secretary, Eastney Modern Boys' School, Reginald Moad, Southsea, Hampshire: R.A.E. course.
Northwood Evening Institute, Potter Street, Northwood Hills. Middlesex: radio amateur examination, morse and practical courses.
Wembley Evening Institute, Copland School, High Road, Wembley, Middlesex: R.A.E. and morse classes.
Wesley Evening Institute: inquiries-"Jeanville". Drighten Road, Addlestone, Werbridge, Surrey: radio and television course.

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# The Practical Wireless OHM'S LAW CALCULATOR 

# A time-saving aid for every enthusiast 

By F. P. Rozee

T1HE relationship between voltage, current, resistance and power is of the utmost importance; a full understanding of the terms and the ability to manipulate them rapidly is the first essential of radio engineering. The relationship is known as Ohm's Law and the importance of this law to the engineer cannot be overstated.

As arithmetical operations are somewhat tedious and mental calculations are unreliable, a mechanical means of computing, with an accuracy better than the tolerance of practical components, becomes necessary.
The Ohm's Law Calculator presented with this month's issue of Practical Wireless has been designed by the author for readers of this publication. It will be found to be a valuable addition to the workshop and a useful companion to the Parallel Resistor Calculator presented with the April 1963 issue of Practical Wireless.

Ohm's Law, using the terms E, I and R, only holds good for d.c. circuits, but it can be used at low frequencies where the load is almost purely resistive.
A mains dropper resistor, although wire-wound and thus to some extent inductive, would, at $50 \mathrm{c} / \mathrm{s}$, behave more or less as if d.c. were being applied. Computations using the calculator may be made at mains frequency for any resistive loads such as fires, heaters, etc.
A word about filament lamps and valve heaters; these will have a very low resistance when cold, only at their working temperature will the resistance agree with the working voltage and current.
The range of voltage, resistance and current encountered in radio and television engineering is far greater than in most other branches of electrical work. Depending on the type of equipment. voltage may range from a few nicrovolts to several kilovolts, current from a few microamps to several
amps and resistance from an ohm or so to several megohms. It is necessary to cater for all of these possibilities and for this reason a range extension chart has been printed on the calculator.

## The Two Scales

An inspection of the calculator shows that it consists of two scales. Scale " A" being the inner or slide and scale " B" the outer. Scale "A" is used for current only and is marked on a logarithmic scale from 1 mA to $1,000 \mathrm{~mA}$, this being the current range that is most likely to be encountered.

Scale " $B$ " is used for resistance, voltage and power ratings. It is marked $1-1,000$, and for voltages would be regarded as 1 to $1,000 \mathrm{~V}$. For resistance, using the left-hand pointer on scale " A", the fixed scale covers from 1 to $1,000 \mathrm{k} \Omega$, while with the right-hand pointer, the range is from 1 to $1,000 \Omega$.

For power rating, using the left-hand pointer, scale " B" covers from 1 to $1,000 \mathrm{~mW}$, and with the right-hand pointer from 1 to $1,000 \mathrm{~W}$. The whole of this latter range is not likely to be required, the largest resistor that would be used is probably a 60 W component, where 200 V are dropped in a 300 mA heater chain. Power dissipations below 100 mW are rarely of interest as a $\frac{1}{10 \mathrm{~W}}$ is the smallest component normally available.

## To Find Voltage or Current

Where the unknown value is that of either voltage or current, set the known resistance against the appropriate pointer. The unknown voltage is then revealed directly above the known current, or the unknown current directly below the known voltage.

## Examples

Known values, $200 \Omega$ and 10 mA . Set ohms pointer to 200 , read 2 V above 10 mA .
Known values, $8 \mathrm{k} \Omega$ and 5 mA . Set kilohms pointer to 8 , read 40 V above 5 mA .
If it is found that with the appropriate pointer on the value of resistance, either the known voltage is not above some part of scale " $A$ " or that the known current is not below some part of scale "B", it will be necessary to use the other pointer and refer to the range extension chart.

## Examples

Known values, $50 \mathrm{k} \Omega$ and 100 mA . Set the kilohms pointer to 50 , it will be seen that 100 mA is not below any part of scale "B". Set the ohms pointer to 50 . We now know that the ohms pointer must indicate in kilohms, the chart (under the column headed "ohms pointer") shows that this could be in association with either microamps and volts, or milliamps and kilovolts. As we have 100 mA , it must be then, that the figure 5 revealed above the 100 is kilovolts. Therefore $50 \mathrm{k}!2$, $100 \mathrm{~mA}=5 \mathrm{kV}$.
Known values, $2 \cdot 5 \mathrm{M} \Omega$ and $200 \mu \mathrm{~A}$. The chart
shows that in association with microamp, megamp could either be set against the ohms pointer and the answer be in kilovolts, or megohm could be set against the kilohms pointer and the answer be in volts. As the first does not allow 200 to appear against scale "B", the second is used. Set kilohms pointer to 2.5 and 500 is revealed above 200. Therefore $2 \cdot 5 \mathrm{M} \Omega, 200 \mu \mathrm{~A}=500 \mathrm{~V}$

## To Find Resistance

Where the unknown value is that of resistance, set the known current in scale " A " directly below the known voltage in scale " $B$ " and read the resistance from whichever of the pointers is on scale " $B$ ". If this is the right-hand pointer the answer is in ohms and if the left-hand pointer, in kilohms.

## Examples:

K nown values, 8 mA and 4 V . See 8 on scale " A" against 4 on scale "B" and read $500 \Omega$ from the ohms pointer.

Known values, 3 mA and 120 V . Set 3 on scale * A" against 120 on scalc " B" and read 40 ks .

Should either or both of the known values not fall in the ranges $1-1,000 \mathrm{~m} \mathrm{~A}$ or $1-1,000 \mathrm{~V}$, refer to the range extension chart, this is quite easy to follow.

According to the top line of the chart, if scale " $A$ " is used for microamps and scale " $\mathbf{B}$ " for microvolts, the resistance reading is in ohms and will be indicated by the kilohms pointer, and so on down the chart to amps and volts where the resistance in ohms is indicated by the kilohms pointer.

Should the ohms pointer be on the scale when microamps and microvolts are associated, or milliamps and millivolts, or amps and volts, the indication would be in thousandths of an ohm. When microamps and kilovolts are associated, the
kilohms pointer would indicate from $1.000 \mathrm{M} \Omega$ up, about equal to an open circuit.

## To Find Power Rating

It is necessary to know both the voltage and the current in order to find the power dissipated in either watts or milliwatts (to accommodate the formulæ $P=F^{2} / R$ and $P=I^{2} R$ would require an extra sliding scale and make the calculator unduly complicated):

Should either voltage or current be the unknown value, finding this using the method described earlier, will take but a moment. The known voltage is set against the left-hand pointer and the power in milliwatts read directly above the known current. If the current is not below some part of scale "B" the power is in excess of IW and it is necessary to set the right-hand pointer to the voltage, when the power in watts is read directly above the known current.

The extension chart includes a multiplying or dividing factor on certain ranges. Where no factor is shown the power dissipation is negligible.

## Resistor Voltage Ratings

The reader is reminded that resistors have a maximum voltage rating. This is not troublesome on the lower resistance values. With a $100 \mathrm{k} \Omega$ component $16 \mathrm{~mA}=160 \mathrm{~V}$ whish is approximately $\frac{1}{4} \mathrm{~W}$. so 160 V is the maximum that would be applied if the power rating were observed. Howcver with a $5 \mathrm{M} \Omega$ component. the full $\frac{1}{5} \mathrm{~W}$ rating is met by 1.1 kV at $210 \mu \mathrm{~A}$, and 1.1 kV is well above the voltage rating for a $\frac{1}{4}$ w carbon resistor.

The following figures will serve as a rough guide to the maximum voltages for carbon composition resistors: $\frac{1}{4} \mathrm{~W}-250 \mathrm{~V}, \frac{1}{6} \mathrm{~W}-400 \mathrm{~V}, \frac{1}{2} \mathrm{~W}-600 \mathrm{~V}$, $1 \mathrm{~W}-800 \mathrm{~V}, 2 \mathrm{~W}-1 \cdot 2 \mathrm{kV}$.

Manufacturer's data should be consulted where possible as the construction, composition and certain techniques in production are among the factors that impose these limits.

## TEST GEAR TECHNIQUES

## -continued from page 545

provision of a connecting terminal for a marker is lacking, the marker signal can be loosely coupled to the wobbulator signal input lead.

## Miscellaneous Points

There are one or two small points to be noted when using an oscilloscope for alignment. Inversion of the trace, vertically, while presenting no real problem. can be confusing. Simply reverse the Y input leads to invert the image. Horizontal reversal is rather more difficult to detect. and requires a frequency shift of the wohbulator input, to determine whether a rise or fall produces the correct lateral shift. Used in conjunction with the marker. this method soon becomes familiar to the operator.

As a final reminder-the response curve displayed is linear, with reference to amplitudeand due allowance must be made when comparing with a published curve, which may be drawn to a logarithmic (decibel) scale.

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## EGG CRATE LINING

SIR.-Recently I was loaned a copy of July 1951
P.W. by a friend. In this issue is an article describing how to make use of papier mâché egg crates as a lining for loudspeaker cabinets of the bass reflex or infinite baffle type.

Having a loudspeaker enclosure which suffered badly from cabinet resonance, I tried this idea and found that the quality of absorption and irregular surface of the egg crate lining allowed 10 W of audio power to be fed into the loudspeaker, without the troublesone "booming" which had occurred previously. I should like to take this opportunity to tharik Mr Simmons who wrote this most useful article.-R. W. Moth (Tettenhall, Staffordshire).

## LANGUAGE BARRIER

S
IR,-I absolutely agree with Mr. Dhenau when he says in his letter (Practical Wireless, July) that Esperanto "would be an untold boon" to wireless communication.

I should like to add that I too know from personal experience that it is much easier to understand and be understood by a foreigner using Esperanto than broken English. Quite apart from the obvious fact that it is a neutral language, one cannot fall into errors and ambiguities of speech. It is impossible to detect the nationality of a reasonably fluent Esperantist, and this surely helps to break down the barriers of language.-Miss E. Storey (Hexham, Northumberland).

## NOVICE LICENCES

SIR,-For some months now I have followed carefully all the correspondence published in P.W. with regard to the licensing of individuals to operate transmitters, and I am convinced that the issue involved is far wider than the question of whether or not "novice licences" should be introduced.

As a nember of a free and democratic society. I believe that every citizen of the U.K. should be able to look upon the use of the ether as his right as an individual and not as the privilege of the initiated few. To refuse anyone permission to use such a convenient method of communication in this day and age is not only narrow-minded but also dictatorial.

Whilst we are always pleased to assist readers with their technical'difficulties, we regret that we are unable to supply diagrams or provide instructions for modifying commercial or surplus equipment. We cannot supply alternative details for receivers described in these pages. WE CANNOT UNDERTAKE TO ANSWER QUERIES OVER THE TELEPHONE. If a postal reply is required a stamped and addressed envelope must be enclosed with the coupon from page iii of the cover.

The Editor does not necessarily agree with the opinions expressed by his correspondents

Of course there must always be limitations on power, frequency, etc., but such things are easily enforced and whereas only limited rights prove acceptable, the complete rejection of these rights is criminal.

No one would deny that. in view of the good name of amateur radio. established over the years, and in consideration of the pioneer work of early hams in "opening up" new bands, the privilege of working certain frequencies should remain exclusively theirs. However, that is no excuse for forbidding the man-in-the-street the use of part of the spectrum and, if we are to be logical about this, we cannot expect every individual to sit and pass a highly technical examination beforehand.R. L. J. Stevenson (Carlisle, Cumberland).

Sir,-I would be grateful if any reader could sell or loon me. . .
the G73. . a manual or any information on the G73 wavemeter.-F. Murdon, 13 Bridge Street, Risca, Newport, Monmouthshire.
. . . the manual of the Bendix RA-1B receiver.-M. J. LaNG, 4 Lynton Road, Burnham-on-Sea, Somerset.
. . . the September 1961 issue of P.W.R. D. Saxton, 165 Gloucester Road, Patchway, Bristol.
the handbook for the oscilloscope No. 11 (A.A. Predictor Mk. 1).-J. E. Griffin, 113 Gladys Avenue, North End, Portsmouth, Hampshire.
. . data or circuit diagram of the R.1392D receiver. - W. Smith, 6 Salisbury Crescent, Blandford Forum, Dorset.

## CORRESPONDENT WANTED

SR,-As a regular reader of Practical Wireless and an enthusiast of amateur radio and television, I would like a correspondent in England of about my own age ( 16 years). I will answer all letters reccived.--Ron Swallow, JNR., I Chauvel Strect, North Ryde, Sydney, N.S.W., Australia.

# BOOK REVIEW 

RADIO AND LINE TRANSMISSION: VOLUME 2
By G. L. Danielson, M.Sc.(Tech.), B.Sc., A.M.I.E.E., and R. S. Walker, Grad.I.E.E., Grad.Brit.I.R.E.; published by Iliffe Books Limited.
295 pages, 224 diagrams, $5 \frac{1}{2} \mathrm{in} . \times 8 \frac{1}{2} \mathrm{in}$. Price 22s. 6 d.

THE whole of the syllabus of the City and Guilds of London Technicians' Certificate examination in Radio and Line Transmission $B$ is covered by this volumc.

Communications students preparing for this third year examination will find that the authors have catered for their requirements using the same well-defined style as in their first volume of this series. As readers of volume 1 will know, both authors are particularly qualified to write on this subject; Mr. Danielson is the Head and Mr. Walker a Lecturer of the Telecommunications Department of the Norwood Technical College.
The range of subjects dealt with is substantially the same as volume 1, but the treatment is deeper: where general introductory statements often sufficed in the carlier book. fuller explanations and mathematical proofs are now given. The text still continues to be mainly of a descriptive nature, and the standard of mathematics required is no higher than ordinary level G.C.E.

For those not familiar with the syllabus of Radio and Line Transmission B it should perhaps be explained that it embraces both general principles and practical aspects of components employed in communications equipment.

## CORRIGENDA

## An Electric Timer

Referring to Fig. 1 (page 131 June issue), the batteries should be transposed: B1 (3V) feeds the buzzer (Z1) and lamp (LPI); B? $(22.5 \mathrm{~V})$ supplies the main circuit.

This article was written by D. Gibson and we apologise for misprinting the author's name.

## Making Tin Plate Chassis

In Fig. 1 (page 73 May issue) tinplate of 16 or 28 s .w.g. was specified in error for 26 or 28 s .w.g. Bearing in mind the simple tools employed. the technique described in this article is applicable only to the thinner gauges of tinplate.

## Double Conversion Communication; Receiver

(1) Referring to Fig. 2a (pages 50. 51 May 1963 issue). L6 should bc redesignated "L6A" and the following item should be added to the components list (page 141 June 1963 issuc):

L6A 2nd oscillator coil (Repanco RO3).
On page 142 (last line but one) the reference to "second oscillator coil $\mathrm{L} 5 "$ should be amended to read "L6A".
(2) The correct type for V13 is VR150/30 (Brimar) or QSI50/40 (G.E.C.).

## Kenilworth Public Address Amplifier

In the components list (page 320 August 1963 issue) the resistors are described as 5 W types; this was a misprint for $\frac{1}{2} \mathrm{~W}$.

## JUST TO PUT YOU IN THE PICTURE

ToO a stcadily increasing band of practical enthusiasts TV no longer signifies passive viewing of BBC or ITV programmes, but a branch of communications offering opportunities for experimental and creative work, just as sound radio has provided for many years past.

Television DX hunting, or the reception of distant TV stations. can be a fascinating hobby where onc's knowledge of v.h.f. and u.h.f. propagation conditions, operating skill, location and-of course-a certain amount of luck! all contribute to the hag of stations reccived. A standard TV receiver can be modified for $\mathbf{D X}$ hunting without much difficulty and there is at the moment a plentiful supply of secondhand receivers on the market at rock bottom prices. A commentary on DX TV now appears regularly in our companion journal Practical Television. This interesting feature lets you know what the other fellow has received and gives guidance on seasonal conditions. and transmitting frequencies and times of operation of the quarry.

Why not build your own TV camera? Starting in this month's Practical Television is a series of articles describing the construction of a closedcircuit TV camera suitable for use with any domestic TV receiver. This camera opens up all kinds of possibilities for home entertainment, as well as more serious applications in picture production. No prior knowledge of camera circuit technique is needed-everything essential is fully explained in these articles.

Of course if you are a little more ambitions there is no reason why you should not build and operate your own TV station. Amateur TV transmissions are permitted on 70 cm and on shorter wavelengths by licensed operators-and there is still an element of pioneering in the establishment of contact and the exchange of pictures with fellow hams on these ultra high freauencies. The background to TV ham activities will be discussed in the November issue of Practical Television and subsequent articles in this new series devoted to Amateur Television Transmission will provide full constructional details for a 70 cm convertor, a comnlete receiver. a transmitter and a camera unit; acrials and ancilliary equioment for the TV ham's station will also be described.

So if you are looking for some further interest in the realm of electronics and communications to help occupy the long Winter evenings that loom ahead. may we suggest that you get into focus and start scanning the pages of Practical Television right away!

Despite our ovening paragraph Practical Television is not ohlivious of the millions of broadcast viewers. In fact many of its pages are devoted regularly to the servicing and maintenance of domestic TV reccivers. Commencing shortly will he a scries describing an all-transistor dual-standard TV receiver for the home constructor.


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Send (preferably) a postal order to cover the cost of the Blueprint (stamps over 6d. unacceptable) to PRACTICAL WIRELESS, Blueprint Dept., George Newnes, Ltd., Tower House, Southampton Street, London W.C.2.

## DOUBLE-SIDED BLUEPRINTS

Each blueprint in this series contains details of two separate instruments or items of equipment.
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$\left.\begin{array}{l}\text { The Berkeley Loudspeaker Enclosure } \\ \text { The Luxembourg Tuner }\end{array}\right\}$ *
$\left.\begin{array}{lll}\text { The PW Troubadour ... } & \text {... } \\ \text { The PW Everest Tuner } & \ldots & \ldots\end{array}\right\}$
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| All-dry Three |  | ... | PW97 |
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The PW Monophonic Electric Organ ..... $8 \%$
The PW Roadfarer * ..... 51.
5). The PT Band III TV converter ..... 116
The Mini-amp * ..... 51.
51. The PT Olympic * ..... 716
The PT Multimeter * ..... 51-
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## SOME EARLIER DESIGNS

T HE following blueprints include some pre-war designs and are kept in circulation for those constructors who wish to make use of old components which they may have in their spares box. The
6/- majority of the components for these receivers are no longer stocked by retailers.
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Midget Short Wave Two ... ... PW38a 216
Simple S.W. One-valver ... ... PW88 216
Pyramid One-valver ... ... ... PW93 216
BBC Special One-valver ... ... AW387 216
A One-valver for America ... ... AW429 216
Short-Wave World Beater ... ... AW436 316
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