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Incorporates TX/RX covernge 2-8 Mels (37.5-150 metresi, and inter com, amplifier. 500 microamp rheck and taning meter, cheul ONLX 65\% Carr. 10


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- Butt in on fortswitch Revolutionars new crystal morro phones with all the teathres of "Mikes" three times th" prife Output level-52db Ummadirect-
ional head. Clipa on of of stameland ional head. Clipa on or off stamelat
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| OZ4 | 579 | $6 \mathrm{CH6}$ | 12'. | 10022 | 27110 | $32 \quad 13 / 6$ | CCH35 | 211. | ECL83 1216 | H23DD | PENA4 1716 | U282 | $22^{\prime}-$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1AS | 61. | 6E5GT | 101. | IOFI | 2612 | 35 A5 151. | CL4 | 1216 | EF9 211. | 1016 | PENB4 1716 | U301 | 2216 |
| ! A7 | 1416 | 6FI | 1516 | 10F3 | 1716 | 35L6GT 1016 | CL33 | $18 / 6$ | EF22 1716 | HL4I 1216 | PEN4DD | U329 | 1716 |
| 1D5 | 141. | $6 F 6$ | 619 | 10F9 | 1216 | $35 \mathrm{W4} 81$. | CYI | 1519 | EF36 716 | HL4IDD | 2216 | $\cup 339$ | 191. |
| ID6 | 10 \% | $6 F 7$ | 151- | 10LD3 | $12 / 6$ | $35 Z 4$ | CY31 | 1519 | EF37 8i6 | $\mathrm{HLA}^{1316}$ | $\begin{array}{ll}\text { PL33 } & 1816\end{array}$ | $\cup 403$ | 1116 |
| f H5 | 1016 | $6 F 12$ $6 F 13$ | 1716 | IOLDII | $15 \%$ | $3525 \quad 916$ | D41 | 126 | EF37A 816 | HL42DD | $\begin{array}{ll}\text { PL38 } & 2319 \\ \text { PL81 } & 1419\end{array}$ | U404 | 101. |
| IL4 | 61. | 6FI4 | 1716 | 10P13 | 211. | 40SUA T51. | D63 | 316 | EF40 151. | HY90 81. | PL82 81. | U4020 | 1516 |
| ILNS | 416 | $6 F 15$ | $14 / 9$ | 10P14 | 201 | 41STH 23'6 | D77 | 516 | EF41 913 | IW $4 / 350101$. | PL83 1016 | UABC8 | $7 \%$ |
| IN5 | 1016 | $6 F 17$ | 121. | 11.03 | 1716 | 42 15\% | D152 | 61. | EF42 1016 | IW4/500 101. | PL820 21\%. | UAF42 | 916 |
| IR5 | 91. | $6 F 33$ | 516 | 11 D 5 | $17 / 6$ | 43 15\% | DAC32 | 1016 | EFSO(E) 316 | KBC32 916 | PM2A 1216 | UB41 | 91. |
| 185 154 | 816 | 6H6 | 216 | 12 A 6 | $6 / 6$ | 5075 15\% | DAF9! | 716 | EF50(A) 41. | KF35 816 | PM2HL 141. | UBC4I | 916 |
| 154 | 816 | 6J5GT | 416 | $12 \mathrm{AH3}$ | $10 \%$ | 50CD6G | DF33 | 1076 | EF80 5 | KL32 1016 | PM22A 1316 | UBF80 | 916 |
| ISS | 91. | 616 | 716 | 12 AT6 | 91. | 32\% | DF91 | 41. | EF85 5/6 | KLL32 1116 | PM24M 2116 | UBF89 | 716 |
| IT4 | 41. | $6 J 7 \mathrm{GT}$ | 916 | $12 \mathrm{AT7}$ | 61. | 50L6GT 9\% | DF92 | 71. | EF86 111. | KL35 916 | PM202 161. | UCC85 | 101. |
| IU5 | 101 | 6K゙7G | 31. | $12 A \cup 7$ | 91. | $618 T \quad 1716$ | DF96 | 916 | EF89 10. | KT2 716 | PY31 16'6 | UCH42 | $8 / 6$ |
| 2021 | 816 | 6K7GT | 1016 | $12 \mathrm{~A} \times 7$ | 916 | 61SPT 1716 | DF97 | 916 | EF91 4i- | KT32 101. | PY $32 \quad 15 / 6$ | UCH8I | 816 |
| $2 \times 2$ | 51. | 6K8GT | $12 / 6$ | 12BA6 | 91. | 62BT 176 | DH63 | 1012 | EF92 51. | KT33C 101. | PY80 81. | UCL82 | 1216 |
| 3 D 6 | 1416 | 6K25 | 1916 | 12 BE 6 | 916 | 75 | DH76 | 716 | EF93 716 | KT36 2816 | PY81 716 | UCL83 | 1316 |
| 3D6 | 146 | 6LI | 1516 | 12 El | 1716 | $77 \quad 12^{\prime} 6$ | DH77 | $8 / 3$ | EF95 151. | KT41 22'6 | PY82 81. | UF41 | 91. |
| $3 \mathrm{Q}^{4}$ | $8 \%$ | 6L6 | 716 | 1217 GT | 916 | $78 \quad 1216$ | DH107 | $13 / 6$ | EL31 1216 | $\begin{array}{ll}\text { KT44 } & 13 \prime 6\end{array}$ | PY83 816 | UF42 | 1116 |
| 3 Q 5 | $10 / 6$ | 6L18 | 1216 | 12K7GT | T 816 | 80 10/. | DH719 | 716 | EL32 51. | KT55 2216 | PZ30 1816 | UF80 | 91. |
| 3S4 | 81. | 6L19 | 211 | 12 KBGT | 11216 | $85 \mathrm{A2}$ 12'6 | DK91 | 91. | EL33 1216 | KT61 18'6 | QP25 14/6 | UF85 | 91. |
| $3 \vee 4$ | 91. | 6L34 | $10 \%$ | 12Q7GT | T 816 | 150821216 | DK92 | 916 | $\begin{array}{ll}\text { EL35 } & 1216\end{array}$ | KT63 816 | QP230 1716 | UF89 | 8'- |
| 5U4G | 416 | 6/30L2 | 1018 | $125 C 7$ | 81. | $15083 \quad 151=$ | DK96 | $10 \%$ | EL37 1816 | KT66 1716 | QP21 1216 | UL41 | 81/ |
| $5 \vee 4$ | 46 | 6LD3 | 916 | $12 \mathrm{SG7}$ | 816 | 185 BT 321. | DL33 | 91. | EL38 2319 | KT71 91- | R10 21\% | UL44 | 2416 |
| $5 Y 3 \mathrm{GT}$ | $8 / 6$ | 6LD20 | 1516 | $125 \mathrm{H7}$ | 41. | 303776 | DL35 | 1216 | EL41 1016 | $k 7741216$ | R19 1916 | UL46 | 211. |
| 5Y3GT | 816 | 6N7GT | 76 | $125 \mathrm{J7}$ | 4\%. | 304716 | DL92 | 816 | EL42 10\% | $K T 76 \quad 12 / 6$ | SD6 8/6 | UL84 | 716 |
| 523 | 101. | 6 MI | 1016 | $125 \mathrm{K7}$ | 81. | 305776 | DL94 | 91. | EL81 1419 | KTIOI 251. | $5 P 4 \quad 14 / 6$ | UU6 | 20111 |
| 524G | 101. | $6 \mathrm{M}^{6}$ | $10^{\prime} 6$ | $12 \mathrm{SQ7}$ | 1116 | 3287716 | DL96 | 916 | EL84 71. | KTW63 716 | $5 P 41 \quad 3 / 6$ | UU7 | 15\% |
| 6 A7 | 1816 | 6P1 | 1716 | 125 N 7 | 1716 | $329 \quad 716$ | EA50 | 21. | EL85 1016 | KTZ41 81. | SP42 12'6 | UU8 | 261. |
| 6 A8 | 101. | ${ }_{6}{ }^{\text {P2 }} 25$ | 1916 | 1223 | 151. | $807 \quad 716$ | EABC80 | 716 | EL90 816 | KTZ63 101 | SP61 3/6 | UY21 | 1516 |
| 6 AB8 | 91. | 6 P 28 | 261. | 13 D 3 | 1216 | 955 41. | EAC91 | 716 | FL91 51. | L63 4/9 | T41 2216 | UY41 | 716 |
| 6AJ8 | 916 | 6Q7GT | $10^{\prime} 6$ | $14 \mathrm{H7}$ | 1216 | 5763176 | EAF42 | 1012 | EM80 10\% | LN309 15/. | TDD 4 17/6 | UY8S, | 616 |
| 6 GK5 | 81. | 6SA7GT | 716 | $14 \mathrm{R7}$ | 1216 | 9002716 | EB41 | 716 | EM81 10\% | LZ319 12'6 | TDDI3C | VP2B | 1716 |
| 6AK8 | 716 | 6SG7 | 716 | 1457 | 211. | 9003716 | E891 | $51 /$ | EYSI 816 | MH4 816 | $17 / 6$ | VP4B | 1716 |
| 6 AL5 | 61. | $65 \mathrm{H7}$ | 616 | 15 A 2 | $17 / 6$ | AC4/PEN | EBC4I | 916 | EY81 1016 | MHD4 1716 | TH41 23/9 | W17 | 816 |
| 6AM5 | 51. | 6SJ7 | 516 | 1502 | 2319 | 25/. | EBF80 | 916 | EY84 1016 | MHL 10'. | TP22 1716 | W76 | 71. |
| 6AM6 | 41. | $65 K 7$ | 716 | 19AQ5 | $10 / 6$ | AC5/PEN | EBF89 | 716 | EY86 916 | MKT4 (5/7) | TP25 1716 | W77 | 51. |
| 6AN5 | 716 | 6SL7GT | 616 | 198G6G |  | $22 / 6$ | EBL2I | 22f= | EY91 91. | 1716 | U14 15/9 | W81 | 61. |
| $6 A Q 5$ | $8 / 3$ | 65N7GT | 516 |  | $24 / 4$ | AC6 21\% | EBL31 | 2176 | EZ35 71- | MS4B 1716 | U16 101. | W142 | 91. |
| $6 A Q 8$ | $9 / 3$ | 6U4GT | 1176 | 2001 | 1216 | ACTP 321. | EC90 | 916 | EZ40 716 | MSP4 1716 | U18/20 101. | W719 | 716 |
| 6AT6 | 813 | 6US | 716 | 20 D 2 | 231. | ACHL 1216 | EC91 | 916 | EZ41 716 | MUl'4 91. | U22 10\% | W727 | 716 |
| 6AU6 | 101. | $6 \cup 7$ | 716 | 20F2 | 2616 | AC/PEN | ECC31 | 101. | EZ80 716 | M×40 17'6 | U24 2916 | $\times 18$ | 1116 |
| $6 \mathrm{B7}$ | 10\% | 6V6G | 51. | 20LI | $26 / 6$ | $17 / 6$ | ECC32 | $101=$ | EZ81 716 | N18 81. | U25 14/= | $\times 61 \mathrm{M}$ | 211. |
| 688 | 41. | 6V6GT | 81. | 20P1 | 261. | ACTHI 3419 | ECC33 | 51. | EZ90 716 | N19 8i. | U26 1216 | $\times 65$ | 2319 |
| 6BA6 | 716 | $6 \times 4$ | 51. | 20P3 | 231. | ACVPI 1716 | ECC34 | 151. | FC2 211. | N37 1816 | U31 916 | $\times 66$ | 211. |
| 6BE6 | 716 | $6 \times 5 \mathrm{GT}$ | 51. | 20P5 | $22 / 6$ | ACVP2 1716 | ECC35 | 81. | $\mathrm{FCl3} \quad 1716$ | N78 1716 | U33 21\% | $\times 78$ | $21 \%$ |
| 6BG6G | 211. | 787 | 81. | 25L6GT | 916 | AC2/PEN | ECC40 | 211. | FCI3C 21. | N108 181. | U35 21\% | $\times 79$ | 211- |
| 6816 | 716 | 7C5 | 81. | $25 Y 5$. | 101. | 2\% | ECC81 | 61. | FWV4,500 | N142 916 | U37 251. | Y61 | 1016 |
| 6BW6 | 71. | $7 \mathrm{C6}$ | $8{ }^{1}=$ | $25 Z 4$ | 916 | $A C 2$ | ECC82 | 916 | $10 \%$ | N147 1816 | $\cup 45 \quad 211$. | Y63 | 91. |
| $6 \mathrm{BW7}$ | 51. | 7D5 | 151. | 2575 | 916 | PENDD21: | ECC83 | 916 | FW 4,800 | N150 10\% | U47 211- | 221 | 1216 |
| $6 \mathrm{BX6}$ | 61. | 7D6 | 151. | 25Z6 | 1016 | $A Z 1 \quad 1516$ | ECC84 | 916 | 10!- | N153 1116 | U50 816 | Z63 | 716 |
| 6 BY 7 | $5 / 6$ | 7D8 | 15\% | 27 SU | 1716 | AZ31 1016 | ECC8S | 81. | GZ30 $\quad 1016$ | N309 1116 | U52 7\%. | Z66 | 1916 |
| 6 C 4 | 616 | 7H7 | 81. | 30 | $13 / 6$ | B36 211. | ECC91 | 516 | GZ32 1116 | N329 101. | U76 716 | $\underline{277}$ | 419 |
| 6C5GT | 81. | 7K7 | 1016 | 30 Cl | $12 / 6$ | B65 8/6 | ECF80 | 1216 | $\begin{array}{ll}\text { GZ34 } & 1316\end{array}$ | N727 716 | U78 7\%. | Z153 | 816 |
| 6C6 | 616 | 7Q7 | 1116 | 30F5 | 1116 | B152 86 | ECF82 | $12 / 6$ | H30 516 | N729 81. | U142 81. | 2719 | 719 |
| 6 C9 | 1216 | $7 \mathrm{R7}$ | $121=$ | 30FLI | 1016 | B309 916 | ECH21 | 22'. | H63 916 | P2 101. | U145 15\% | ZD152 | 916 |
| 6 Cl 10 | $12 / 6$ | 7\$7 | $10^{\prime} 6$ | 30LI | 1116 | 8329916 | ECH35 | 21\% | HBC90 916 | PCC84 91. | U147 71. |  |  |
| 6CD66 | $27 / 6$ | 7 Y 4 | 716 | 33 P 4 | $22 / 8$ | $8339 \quad 916$ | ECH42 | 10\% | HL92 616 | PCF80 916 | U153 916 |  |  |
| 6 61 | 81. | 803 | 41. 419 | 30 Pl 12 | 1116 | B719 97 <br> 817  | ECH8! | 91. | HLI330D | PCF82 816 | U191 2016 |  |  |
| 602 603 | 51. | 98W6 | 1419 | 30 Pl 16 | 101 151 | CBLI 1716 | ECH93 | 1216 | H23 1016 | PCL82 101. | U251 1716 |  |  |
| 6 D 3 | 151. | $\cdots \mathrm{IOCI}$ | 181 | 30PLI | $15 \%$ | CBL3! 21\%. | ECL80 | 91. | $\begin{array}{ll}\text { HL23 } & 1216\end{array}$ | $\begin{array}{ll}\text { PCL83 } & 12 \prime 6\end{array}$ | U281 20'. |  |  |


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| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| RMI | 61. | I8RA | 1-1-8-1 | $4 / 6$ | 16RE 2-1-8-1 | 816 |
| RM2 | $8 \%$ | I8RA | 1-1.16-1 | 616 | I8RA 1-2-8-1 | 111. |
| RM3 | 91. | IGRA | 1-1-16-1 | 816 | 14A86 | 171. |
| RM4 | 1616 | 14RA | 1-2-8-2 | 181 | 14 A 97 | 2316 |
| RM5 | $22 \%$. | I4RA | 1-2-8-2 | $21^{\prime}$ | 14 Al 100 | 241. |

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| OA2 | 1716 | 6F6G | 1713 | 10F9 |  | $\begin{aligned} & 3574 G T \\ & 3575 G T \end{aligned}$ |  |  | $7$ |  | $\begin{aligned} & 517 \\ & 9 I_{1} \end{aligned}$ |  | $\begin{aligned} & 2477 \\ & 20^{\prime} \end{aligned}$ | PL820 | $\begin{aligned} & 12 \prime 8 \\ & 18,8 \end{aligned}$ |  | $916$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| O82 | 17 | 6F11 | $17 / 3$ | $10 F$ | $11$ | ${ }_{43}^{35 \mathrm{Z} 5 \mathrm{GT}}$ |  | DH77 DK32 |  | EF4 <br> EF42 |  | HVR2 |  |  |  |  |  |  |  |
| OZ4 |  | 6F12 | 115 |  | (5) | 50 | $10 \%$ |  | 616 | EF50( | ( 71 | KF35 | 816 | PX4 | 1016 | UBC41 | 816 | ${ }_{\text {Y63 }}$ | 16 |
|  | 121. | 6F15 | $5 / 3$ | 10 PI 4 | 1913 |  |  | DK92 |  |  | ) 5 |  | 816 |  | 1617 | UBCBI | $11 /$ | Z63 | 16 |
| C | 1216 | 6 F23 | 1016 | 1246 |  |  |  | DK96 | 16 |  | 5 | KLL3 | 2417 | PY | 216 |  |  |  |  |
|  |  |  |  |  |  |  | 16 | DL33 |  |  | 1016 |  |  | PY80 | 16 | UBF89 | 916 | 277 |  |
| G6 | 171 | 6F32 | 101 |  | , |  | 11 | DL6 | 1716 | EF | , |  | $10^{\prime}$ | PY8 | 816 | UBL21 | 23/3 | 2719 | b. |
| H5G | 1016 | 6F33 | 716 |  |  | - |  | DL6 | 15. |  |  | KT36 2 | 碗 | PY82 |  |  |  |  |  |
| 114 | $3 /$ |  | 616 |  | 81 | 78 | 616 | DL | 15. | EF8 | 1016 | KT4 | $23 / 3$ | PY83 | 816 | UCC8 | 91. |  |  |
|  |  |  |  |  | 1216 | 80 |  |  |  | EF | 91. | KT4 | 1216 | PY88 | 13/3 | UCF80 | 1617 |  |  |
| ILN | 51 | 615 | 5. | 12AT6 | 716 | 83 | $15 \%$ | DL9 | 716 | EF9 | 416 | KT61 | $12 / 6$ | PZ30 | 19111 | UCH2 | 23/3 |  |  |
|  |  | 61 | 516 | 12AT7 | 61. | 85A |  | DL9 | 816 | EF92 | 416 |  | 71. | O |  | $\cup \mathrm{CH} 4$ | 916 | CG4E | 16 |
|  | 616 |  |  |  | 23/3 | AC |  | DM | 716 | EF9 | 1313 | KT6 | 15. | QP25 | 1416 | UCH | 916 |  | 16 |
| 154 | 110 | 677 GT | 1016 | $12 A U$ | 616 | 90 | 671 | E80 | 30\% | EF9 | 1313 | KT8 | 241. |  |  | UCL | $11 / 6$ | CG |  |
| 155 | 6 |  |  | 12 | 12'8 | 90 C | 161. | E83F | $30 \%$ | EF183 | 1817 | KTW61 |  |  | $10 \%$ | UCL8 | 1913 | C |  |
| IT | 316 | 6K7GT | 61. | 12AX | 716 | 90 C | 3716 | EA50 | ${ }^{2 \prime}$ | EF18 | 1216 | KTW62 | 16 |  |  | UF41 | 16 |  |  |
| 10 | 61. | 6 K 8 GT | 016 | 12BA6 |  | 101 | 1316 | EA76 | 916 | EK32 | 816 | KTW63 |  | R18 |  | UF42 | $12 / 6$ | 6. 8 |  |
| 2 D 21 | 51. |  | 616 | 12 BE 6 | 91. | 1508 | 181. | EABC80 | 9\% | EL32 | 5 | KTZ41 | $8{ }^{8}$ - | R 19 | 19111 | UF80 | 1016 | A70 | - |
| 2 P | 2616 | 6 K | 1911 | 12 | $21 / 3$ | 161 | 1016 | EAC91 | 416 | EL33 | 216 | KTZ63 | 716 | RGI |  | UF8 |  | A73 | 1-1 |
| 2 | 416 | 6 LI | $23 / 3$ | 12 | 301 | 185 | 312 | EAF42 |  | EL3 | , |  |  |  |  | UF | 11 | OA79 |  |
| $3{ }^{2} 4$ |  | 6L6G | 81. | 12 J | 416 | 304 | 016 |  | 216 | EL38 | 2616 | MHL | 716 | RK | 716 | UF89 | 91. | OA81 | . |
| 3 A | 1016 | 6L6M | 916 |  |  |  | 1016 | EB41 | 816 | EL41 | 1. | MHL |  |  | 2216 | UL41 | 91. | OA86 |  |
|  |  | GT | 716 | 2 K 5 | 11 | 807 |  | EB91 | 41 | EL42 | 1016 | ML | 816 | SP |  | UL44 | 2616 | OA91 | 16 |
| 3 D 6 | 51. | 6L18 | $13 \%$ | $12 \mathrm{K7}$ | 16 | 956 | 1. | EBC3 | 23/3 | EL81 | 617 | MS | $23 / 3$ | SP |  | UL | 1416 |  | 16 |
| O | 716 | 6 L 19 | 313 |  | $4{ }^{\prime}$ | 1821 | 67 | EBC | 51. | EL83 | 11 | MU1 |  | ${ }^{\text {SP4 }} 4$ | 1216 | UL84 | 816 | OA210 |  |
| 3 O 5 | 916 | 6LD2 | 11 | 12 C |  | 4033 L | 216 |  | 816 | EL84 | 716 |  | 23'3 |  |  | UM4 | $17 / 3$ | OA211 |  |
| 35 |  |  |  | 12SA7 | 816 | 5763 | 1216 | EBC |  | EL85 | 1311 | N10 | 19,11 | SU2 |  | UM34 |  | 19 |  |
| 3 V 4 | 716 | 6 P | 1911 | $125 C 7$ | 816 | 7193 |  | EBF |  | EL86 | 1713 | N108 | 23/3 |  |  | UM80 |  | OC19 |  |
| $5 R 4 G$ | 716 |  | 2618 | 1254 |  |  | 516 |  |  | EL9 |  |  |  |  |  |  |  | OC23 |  |
|  |  |  | ${ }^{616}$ | ${ }_{12512}^{1257}$ | 6 | $9002$ | 516 | EBF8 | 2916 | EL820 | 1817 18 |  | 316 | TH233 | 3312 | U47 |  | OC28 | 251 |
|  | 16 |  | 101. | 125k7 | 61. | 5-pin | /3 | EBL2 | 23/3 | EL822 | 191. | 80 |  | TP22 | 151- | U 48 | 2616 | - | 2516 |
| $5 \mathrm{Z3}$, | 11 |  | 16 | 12SQ7 | 1116 | pi |  | EBL31 | 23/3 | EM34 | 916 |  | 1311 | TP25 | 15' | UU9 |  | OC |  |
| 524 |  |  | 716 | R7 |  | PE |  | EC52 | 16 | EM7 | $23 / 3$ | PC |  | TP26 | 33/2 | UYIN | 187 | OC | $1{ }^{1}$ |
| 6 A7 | 1016 | 670 |  |  | 1016 |  |  | EC54 | ${ }^{6} 1$ | EM80 |  | PCC85 | 916 |  |  | UY21 |  |  |  |
| AB |  | 6SH7GT | 81 | 57 | 27110 | SPE | 716 | EC70 | 1216 | EM81 |  | PCC88 | 181. | U12/ |  | UY4t |  | OC6 | 16 |
| 6 AC | 41. | 6S17 | 81 | 19 AQ | 1016 | ${ }^{\text {AC/ }}$ A ${ }^{\text {a }}$ | 33'2 | EC92 | $13 / 3$ 516 | EM8 |  |  |  |  |  |  |  |  |  |
| ${ }^{\text {a }}$ | 516 |  | $6 \cdot$ | 19 HI | 10\%. | ATP4 |  | ECC32 | $\begin{aligned} & 516 \\ & 816 \end{aligned}$ | EM8 | $17 / 3$ | PCF82 | 1016 |  |  | VMS4B |  | OC72 | I6. |
| AG | 716 |  |  |  |  | AZ | $10 \%$ | ECC3 | 2477 | EY51 | 53. | PCFE | 167 | 12 | \%. | - | 1216 |  |  |
|  | 4 |  | T 9 |  | 2616 | AZ41 | 1311 | ECC35 | 816 | EY83 | 1617 | PCF86 | 15\% | U24 | 29110 | VP4 | 151. | C |  |
| 6 6 | 46 | 6S57 | 81. | 208 | 2616 | B36 | 151. |  | 23/3 | EY8 |  |  |  | 25 | 1711 | VP2B | 1416 | C77 |  |
|  | 716 |  | 216 |  |  | BL6 | 716 | ECC81 | 61. | EY86 | 91. | PCL83 | 1018 | 426 | 10 | $\checkmark \mathrm{VP}$ | 23/3 | OC78 |  |
|  | 71. | 6 U | 716 |  | 2616 | - | 1216 | ECC82 | ${ }^{16}$ | EZ35 | 61. | PCL84 | 1216 | 431 | 916 | VPI3C | 71. | C81 |  |
| 6AU | 101 | 6 U | 816 |  | 23/3 | CIC | 1216 | ECC8 | 716 | EZ40 | 71. | PCLB | 16 | 43 | 2616 | VP | 616 |  |  |
| 6A | 12 |  | 71. | 25 | 1016 | CBL | 2616 | EC |  | EZ41 | 71. |  | 1617 | $\cup 35$ | 2616 |  | 61. | C171 |  |
| 688 |  | 6 V | G 81 |  | 10'. | CBL | $23 / 3$ | C8 | 816 | EZ80 | 1. | PEN | 23/3 | 37 |  | RR | 1. | OC200 |  |
| 68A6 | 716 |  | 51. |  |  |  | 23/3 | ECC88 | 181. | EZ81 |  | PEN |  | U43 |  |  |  |  |  |
| 6BE | $6 \cdot$ | 6×50 | 61. | 25 | 916 | CK506 | 616 | ECC91 | 516 | $\mathrm{FWW}_{4 / 50}$ |  | PEN4D |  | $U 45$ 450 |  | VT61 |  | OCP71 |  |
| 68 G | 23/3 | 6/30L2 | 101 | 52 | 916 | CL33 | 1913 |  |  |  |  |  |  | U5 |  | VT50 |  |  |  |
|  | 8'. | 7A7 | 1216 $21 / 3$ |  |  |  | 1816 | ECF82 | 10 | FW4/8 GU50 |  | PEN25 |  | 454 | 19111 | W8iM | $6 \%$ | T13 |  |
| $68 Q 7$ | 15\% | 787 | 816 | 28 D 7 | 1 |  | $11 \%$ | ECH | 2616 | GZ30 |  |  | ', | U76 | 61. | W107 | 1817 | TPI |  |
| 6BR7 | 1216 | C5 | 81. | 30 Cl | 81. | D | 3'. | ECH2 | 23/3 | GZ32 | 10\%. | PEN4 | 2616 | 48 | 51. | W729 | 1911 | TP2 |  |
|  | 25 |  | 81. |  | 研 |  | $10 / 6$ |  | 6.6 | Gz33 | 1911 | PEN 4 | 1916 | U107 | 1617 | $\times 24 \mathrm{M}$ | 247 | TSI |  |
| 68W |  | , | \% | 30FL | sor |  | 41. | $\mathrm{ECH}_{4}$ |  | GZ3 | 1 | PEN | 16 | 4191 | 1617 |  | 15'- | TS2 |  |
| 6BW | 61 | 7R7 | 1216 | 30 L | 81 | DaC32 | 1016 | ECH81 | 91. | GZ37 | 19111 | PEN 383 | 23/3 | 4201 | 161 | $\times 61$ (C) | 16 | TS3 |  |
| C4 | 5\% | S7 | 916 | 30 L 15 | $11 / 6$ | DA | \% | ECH83 | 13111 | H63 | 12'6 | PEN453 |  | 4251 | 141. | $\times 63$ | 91. | TS |  |
| 6 CS | 616 | V7 | 816 |  | 121 |  | 816 |  | ${ }^{91} 16$ |  |  |  | 33/2 | U281 | 2111 |  | 1216 |  |  |
| $6{ }_{6} 6$ | 616 | $7{ }^{\text {r }}$ | 76 | P12 | 716 | DD41 | 13111 | ECL82 | 1016 |  |  |  |  | U282 | $22 / 7$ |  | 1216 | $\times \mathrm{AlOI}$ |  |
| 6 C 9 | 1316 | $8{ }^{80} 2$ | 36 | 30 P 19 | $12 \%$ | DET25 | 76 | ECL83 | 1913 | HL2 |  |  |  | U301 |  |  |  |  |  |
| 6 Cl 10 | 9 | $8{ }^{88}$ | 416 | 30PL1 | 1016 1616 | DF33 | $\xrightarrow{1016} 1$ | ECL8 <br> EF | 1617 $23 / 3$ | $\begin{aligned} & \mathrm{HL2} \\ & \mathrm{HL2} \end{aligned}$ |  | $\begin{aligned} & \text { PL33 } \\ & \text { PL36 } \end{aligned}$ | 1913 | U329 |  |  | $23 / 3$ $23 / 3$ | $\begin{array}{\|l\|l\|} \text { XA } 103 \\ \text { XA } 104 \end{array}$ |  |
| 6 CD | 36 | 82 | $15 / 3$ | 30PL13 | $21^{1 / 3}$ |  |  | EF9 |  |  | 16 | $\mathrm{PL}$ |  | U339 |  | $x$ | $23 / 3$ 1713 | $\times$ Al04 |  |
|  | 616 | $\begin{aligned} & 9 D 2 \\ & 10 \mathrm{Cl} \end{aligned}$ | 1315 | ${ }^{\text {35A5 }}$ 356T | $21 / 3$ $9 / 6$ |  | $\begin{aligned} & 3 / 6 \\ & 816 \end{aligned}$ | $\begin{aligned} & \text { EF22 } \\ & \text { EF36 } \end{aligned}$ |  |  |  | PL81 |  | U404 | 816 | XD(1.5) | 616 | XB103 |  |
|  | 6 | 10 C 2 | 2616 |  |  |  |  | EF37A | 81. |  |  |  | 16 | U801 2 | 29110 |  | 181 | B104 |  |
|  |  | 10 | 121. | 35 | 1016 | DH63 | 616 | EF39 | 5.6 |  |  | 83 | 9 | 020 |  | XFYI2 | 9 | 01 |  |

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119
19 16 16 mid 450
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A.C.ID POCXET MULTI-METER KIT , Hn movite coil meter scule A.C.D.C. POGAC M Late $0-500-1000.550 .0-501$ wiliamps $0-10,0-100$ Ohms ratge 0-10.00. Front panel tullge switch wirewound pot (ior ohte zero setting) wogle Front reslstor and rectifer. 19/B. P. \& P. $1 / B$ Wiriug dhasram i/, iree with bit.
 modified as an aserial converter (instructions supplied). 32/6, plus 3/6 $\mathbf{P}$. \& $\mathbf{P}$. HEATER TRANSFORMER to snit alove, $200-260 \mathrm{v}, 6 / \mathrm{f}, \mathrm{phls} 1 / 6 \mathrm{P}$. \& $P$. MAINS TRANSFORMERS. All with tapped primaries, $200-250$ wolts, $0-160$

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REALISM AT INCREDIBLY LOW COET，CAN IBE ASSEMBILSD IN HALE AN HOUR The Recorder incorporates the Latest Collaro Studio Tape Transcriptor，The Linear LT45X High Best quality Tape listed $2 / 6$ ，and a Handsome Portable carrying Cabinet with latest attractive two Best qualicy iape listed $2 / 6$ ，and a 1 andsome Portable carryor high．IIsted 4.10 .0 ，and circuit．Total cost if purchased individually approximetely f40．Performance equal to units in the $\AA 50$－ 880 class．S．A．E for leaflet．

## HICH FIDELITY 12－14 WATT AMPLIFIER TYPE A11

PUSH－PULL ULTRA LINEAR OUTPUT＂BUILT－IN＂TONE CONTROL PRE－AMP STAGES Two thput sookets with assochated gram，es in A10，Eifg sensitivity． Inoludes 5 Falves HCOB ，ECC83， LeA4，RLDA，5Y3．High Quallty sec－ thonally wound output trangformer specially desifned for Ultra Linear
operation and reliable small con－ operation，and reliable small con－ densers of current manufgoture．IN－ DIVID TAL CONTROLS FOR BASS AND TREBLAEA response +30 a $30-30000$ alcs Six negative ieadoack loops fum level 60 D．B dow On loops． milivolts INPUT＇required for FULL
 OUTPUT．Suitable for use with all makes and types of dick－ups and microphones．Com－ parable with the very best designs．For STANDARD or LONG PLAYMAGRECGRDS socKET with plug provides 300 ． 30 mA ．and $6.3 \forall .1 .5$ a．For supply of a FADIO PEEDER UNIL．Size approx． $129-71 \mathrm{I}$ ．For A．C．mains $200-250 \mathrm{v}, 50 \mathrm{c}$ ．p．s．Output for 3 and 15 ohms speakerg，Kit is completo to last nut．Chassis is fully punched．Full instructions and point－to－point wiring diagrame supplied．Only if fins Carr． （以及）

8 Gns．
If required louvred metal cover with 2 carrying handles can be suppiled for 18／9，TERMS DN ASSEMBLED UNITS．DEPOSIT $84 / 9$ ，and 8 monthly payments of $24 / 9$ ．Send 3．A．in．for lilustrated leafiet detalling Ready－to－assemble Cabigets，Speakers．Micro－ phones，otc．，with cash and credit terms．

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A complete set of parts tor the construction of a stereo－ phonic amplifier giving 5 watts high quality output on volts，suitable for all crystal stereo heads．Ganged Bass and Treble Controls give equal variation of＂ljft＂ and＂cut＂．Provision is made for use as straight （monaural） 10 watt amplifler．Valve line－up ECC83， ECC83．EL84，ELZ81．Outputs ior $2-3$ ohm speakers． Point－to－Polnt wiring diagrams and in－ 8 Ens． Full constructionsi details and price list 2／6．Carr．10／－

## $25^{\frac{1}{3}}$

GNS．
Carr．
17／6


H．P．TERMS．Deposit 85.7 .6 and 13 monthly payments of 2 gns．Cash price it settied in 3 months．

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price e8．19．6．Carr． $5 / 6$ ．
COLLARO RC 457 －SPEED MIXER AUTO－CHANGERS．Turnover Studio Pick－up head．for $200-250$ v．A．C．27．19．6． Carr．4／6．
TIE SKYFOUR T．R．F．RECEIVER A design of a 3－valve long and Medium selentum rectifler High catp $\mathbf{H}$ Ftase and low distortion detector $V$ aive interso $6 \mathrm{K7}$ SP61．6V8G．Selectivity and quailty excelient．Simple to construct．Point－to－ Point wiring diagrams，instructions and parts list，1／9，maximum bullding costs e4．19．6，inc．Attractive Walnut veneered wood cabinet $12 \times 6+5 \mathrm{fin}$ ．
1）．C．SUPPLY KIT． 12 マ． 1 a．consiating of a partially drilled metal case，mains trans．．F．W．Bridse Rectifler． 2 fuseholders and tuses．Change Direction switch，vari－ $200-250$ peod regulator and circunt． Trains．Limited number available at $33 / 9$.

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All for A．C．Mains 200－250va， 50 ecs．
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MICROPHONE INSERTS，OTystal type $6 / 9$ ．


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 or 12 v． 4 amps．Fitted Ammeter and variable charge rate variabler charge rate selector．Also selec－ 12 p charging ．Ou vrod steel case with stoved blue hammer finished．Fused and ready for 69／6 use with Carr．5／－ mains and output leads．Terms leads． monthiy payments 13／3．

ASSEMBLED
6 จ．or 12 ． 2 amps．
Fitted Arnmeter and selector ${ }_{12}$ plug ror $\nabla$ ．or merai case merai case fin－ shed attractive
hammer blue． Roady for use With maine and outonit leads．Double Fused．Only Carr．3／9．49／9

BATEERY CHARGER KITS Consisting of Mains Trans－ Rectifler，well ventilated stoel case．Fuses，Fuse－holders． Grommets，panels and drcuit． Carr． $3 / 6$ extra． 6 v．or 12 v． 1 amp．
As above，with A．．．．．．．．．．．． $24 / 9$
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2519 $6 \mathrm{\nabla}$ ．or 12 v .2 amps ．．．．．．．．．．31／8 8 ．or 12 v． 2 amps．inclu． ． $48 / 9$ Sive ot Ammeter ．．．．．．．．．42／9 5
 6 ．or 12 V． 4 amps． Amirheter and varlable charge rate selector ．．．．．． CHARGER AMMETERT，


R．S．C．MAINS TRANSFORMERS（GUULLY ${ }_{\text {GAN }}$ ）
Interleaved and inupreraated．Prim－
 $250-0-250$ v． $70 \mathrm{~mA}, 2.3$ v． $2 \mathrm{a}, 5 \mathrm{v} .2 \mathrm{a}$ ．． $17 / 9$ $850-0-350$ v． $80 \mathrm{~mA}, 6.3$ v． 2 a． 58.2 a．． $19 / 9$ $250-0-250$ 下． $100 \mathrm{~mA}, 8.3$ v． 2 A． 8.9 ₹． 1 a $19 / 9$

 $100-0-300$ v． $130 \mathrm{~mA}, 6.3$ v． $4 \mathrm{a}, 6.3$ v． 1 a for Mullard 510 Amplifler

## ． 8919



 $250-0-360$ \％． $150 \mathrm{~mA}, 3.3$ v． $4 \mathrm{a}, 5$ \％． 3 a． 2949
 Midset type $21-3-31 n$ ．
 $300-0-300$ v， $120 \mathrm{~mA}, 6.3$ v． $4 \mathrm{~m}, 5 \mathrm{~F} .3 \mathrm{a} \ldots 9 \mathrm{~F} / 9$
 $350-0-350$ v． 150 mA .8 .3 v． 4 a． 5 v．3 a．． $35 / 9$ 6.3 V． 4 ®．C．T．． 5 ष．3

49／9

FLLAMEUNT TRANSFGRMERS
All with $200-260 \mathrm{v}, 50 \mathrm{c} / \mathrm{s}$ ，primaries 8.3 y ． 1.5 a， $5 / 8: 6.3$ v． 2 a．7／6： $0-46.5 \nabla .2$ a．7／8； 12 จ． 1 a．7／11： 6.3 จ． 3 a， $8 / 11 ; 6.3$ จ． 8 a，
$17 / 6$ ： 12 v． 1.5 a twice． $17 / 6$.

OUTPUT TRANSFORMERS
Midget Battery Pentode 68：1 for 3S4．etc．

819
$\therefore .8 / 8$
.819 Smali Pentodie． 50000 tö3a： Smail Pentode 78，000 0 to 3 n $\because \quad 3 / 9$
$\because \quad 51$ Standard Pentode 5.0000 to 30 Standard Pentode 7／8．0000 to $3 \Omega$ 10,1000 to 30
Push－Pul $10-12$ wätts $6 \dot{\mathrm{~V}} 8$ to 3 a or
 Push－Pull 12－12 watts to match 8V6 $18 / 8$ Push－Pull ELB4 co 3 or $15 \Omega$ © $\quad \rightarrow 18 / 8$ Pueh－Pull for Mullard 510 Ultra Push－Pull on watts．soctionalis wound 5 L 6 ．KT68 etc．，to 3 to 150 ．， $49 / 9$

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$150 \mathrm{~mA}, 710 \mathrm{H} 250$ ohms．
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$80 \mathrm{~mA}, 10 \mathrm{H}, 10 \mathrm{~F} 400 \mathrm{ohms}$ ．


## CHARGER THANSFORMERS

All with $200-230-250$ v． 50 c／s Primaries： $0-9-15$ v． 11 a，11／9：0－8－15 v． 2 a．14／9：0－9－15 v．3 a．16／9：0－9－15 v．5 a，19／9：0－0－15 $\vee$ ． a．28／6：0－9－15 v．8 a．28／9．

AUTO（Step up／Step down）TRANS． 0－110／120－230／250 \％． $50-80$ watts． $13 / 9:$
atw，2718．
IICROPHONE TRANSFORMERA 120：1 high srade，clamped．6／9：120：1 Potted．Mu－metel screened． $9 / 9$ ．

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transformer is used with gof output transformer is used with 807 output Falves, All components are chosen for reliablity, S1x vaives are used EF86. Eass, and Treble Cont are Separate Bass and Treble Controls are provided. Minimum input required for full output os oniy 12 millivolts so that ANY KiND SUITABLE. The unit is destrned for SLIABBLE. The unit is designed for
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| $\mathrm{HF}^{\text {¢ }}$ 5 | $7 \%$ | PL36 | 14／8 | $\times 109$ | $15 / 9$ | 2574 | H16 |
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| ELS | $11 / 3$ | PY80 | 716 | $1 \mathrm{A5GT}$ | 81－ | 48 Ul | 101－ |
| ELat | 51\％ | PY81 | 9／8 | 1A7GT | 12／－ | 77 | 8／－ |
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[^2]REDUCTIONS IN MULLARD TRANSISTORS OC19 481-|OC72 81-|OC78D 81-

 \begin{tabular}{cc|cc|cc|cc|}
\hline OC44 \& 111. \& $0 C 73$ \& 161. \& $O C 84$ \& $11 / 6$ <br>
$O C 45$ \& 101. \& $0 C 75$ \& 81. \& $O C 170$ \& 1316

 

OC45 \& $101-$ \& OC75 \& 81. \& OCC171 \& 1416
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# Practical Wireless 

VOL. XXXVII No. 655 SEPTEMBER, 1961


## MORE FREE BLUEPRINTS

INCLUDED with each copy of our next issue, will be a blueprint for our new receiver-the P.W. Tutor. As its name implies, it is designed essentially for the beginner, although it will also be found instructive for those who have not built a radio set for some time. The great success of our previous series for the beginner-The Beginner's Constructional Course and others-shows that there is a real need for this type of apparatus. Young people are continually leaving school and entering the radio field, and wish to add to any knowledge they may have gained at school, with practical work, which at the same time, will introduce them to new aspects of construction.
It is natural that the beginner will need to start from the very simplest point which means the crystal receiver, but the transistor must also be employed as this is becoming more and more to be accepted in place of the valve. Printed circuits are, of course, very widely used to-day, but it was felt that it would be rather too involved to try and use these in a receiver of the progressive kind, so the technique of utilising standard tag-boards has been adopted, and the receiver is built on these. The constructor goes from stage to stage, gradually adding parts and so advancing the design of each set.

## NOVEMBER AND DECEMBER BLUEPRINTS

In the November and December issues, further free blueprints will enable more additions to be made to the preceding receivers. Included in the second and third blueprints will be a transistor battery superhet tuner together with its amplifier, which will also be suitable for a battery operated record player.

The blueprints will be such that the stages of construction are complete in themselves and each reader will be able to choose a design to suit his pocket and requirements.

## INCREASED PRICE

The steadily increasing costs of production, paper, and other materials have made it necessary to increase the price of this magazine from the October issue-or to reduce it in size. We know that the latter procedure would not meet with approval and that most readers would rather pay an increased price than buy a smaller edition of Practical Wireless.
We shall, of course, keep up the same high standard of articles and endeavour to make each issue appeal to the widest possible circle of readers-catering both for the absolute beginner, as in the blueprint mentioned in the above paragraphs, as well as the advanced constructor and amateur transmitter. Blueprints in later issues will be designed for more experienced readers and we shall continue to provide the maximum service not only to individuals, but also to clubs and societies. From the October issue. therefore, the price of Practical Wireless will be increased to $2 /$-.

The October issue of Practical Wireless is published on September 7 th-order your copy now and be sure of obtaining the first of the free blueprints.
 Our next issue, dated October, will be published on September 7th

## Ronned the Worlal of Wireless

## POTENTIAL AND CURRENT NEWS

## Broadcast Receiving Licences

THE following statement shows the approximate number of Broadcast Receiving Licences in force at the end of May, 1961, in respect of wireless receiving stations situated within the various Postal Regions of England, Wales, Scotland and Northern Ireland. The numbers include Licences issued to blind persons without payment.

| Region |  |  |  | Total |
| :---: | :---: | :---: | :---: | :---: |
| London Postal . . |  | - | -• | 687.207 |
| Home Counties.. | $\because$ | $\because$ | $\because$ | 649,332 |
| North Eastern ${ }^{\text {M }}$ | . | - |  | 469.160 |
| North Western ${ }^{\text {. }}$ | .. | $\because$ | $\because$ | 511,084 |
| South Western |  |  |  | 388.260 |
| Wales and Border | Countres |  | $\because$ | 227,571 |
| Total England and | Wales | - | - | 3.372,460) |
| Northarn ir ${ }^{\text {a }}$ |  | -. | . | 376,605 |
| Northern Ireland | - |  |  | 117,560 |
| Grand Total | -* | -* |  | 3,866,625 |

## Tape Recorder Company Interests Change Hands

A wholly-owned subsidiary of Mulficore Solders Limited, Multinusic Limited. announces that it has disposed of all its interests in Reflectograph Tape Recorders to Pamphonic Reproduccrs Limited, a Pye Group Company. Pamphonic are now responsible for all service of Reflectograph recorders. Enquiries from the public concerning Reflectograph recorders should now be directed to Pamphonic Reproducers Limited at 17 StratIon Street, London, W.1.

## Expansion at Valve and Capacitor Factory

A FURTHER expansion programme for the production and marketing of new and present types of valves and capacitors has recently been started by Standard Telephones and Cables Limited.

Manufacture of both these nroduct lines is at the modern S.T.C. factory in Paignton, Devon. In the Valve Division plant, the principal products are valves for use in telecommunication systems, such as microwave radio links and transoccanic telephone cables.


Sealing the glass envelope of a water-cooled power triode, at the S.T.C. factory at Paignton, Devon.

Expansion is envisaged in the production and application of many types of valve.

Capacitors, like valves, have been key products of S.T.C. for many years and were formerly made mainly at North Woolwich, London, with a satellite plant at Treforest in Wales. The success of the Valve Division operations Ied to the building of a sccond factory for the manufacture of capacitors.

## Multi-channel Radio Telephone System for British Railways

AS part of the British Railways modernisation programme, its north-eastern region is to be equipped with a high-frequency multi-channel radio telephone system betwcen York and Newcastle, via Darlington--a distance of 78 miles. This will be the first such installation to be used by British Railways.
The radio cquipment is to be supplied and installed by Marconi's Wireless Telegraph Co. Ltd., and the carrier equipment by Automatic Telephone and Electric Co. Lid. The system is designed for a maximum capacity of 300 telephone channels, but initially 180 will be in operation. It will handle the telephone traffic between the regional centres, with full facilities for through subscriber trunk dialling.

The installation is duplicated to operate on a main-standby basis, with automatic changeover in the event of failure. In this connection, particular features of the equipment are the relatively low number of valves used and the simplicity of the R.F. transmission circuits, which employ one microwave klystron per unit equipnent. both factors which greatly reduce the probability of cquipment failure. The telcphone channelling equipment is of the latest design. using transistors throughout and printed wiring cards for mounting all components.

## Radio and Television Exhibition

THE Radio and Television Exhibition of France will be held this year from Scptember 14th to 26th in Paris-Porte de Versailles. It will be situated in the Parc des Expositions-Halls 101 to 110. This exhibition is undertaken by the Radiodiffusion Télćvision Francaise and by the Fédération Nationale des Industries Electroniques.
The exhibition, which covers all fields of radio and television, is open to all foreign visitors but is for French exhibitors only. The utilization of the "Palais des Sports" cnables the organizers to receive the largest audience in conditions of comfort. A studio
with 1,000 seats for broadcasting lyrical and dramatic programmes, and high fidelity demonstrations will be included in the exhibition.

## Minicabs Radiotelephone System

WITH the opening of the London minicab scheme, Pye Telecommunications put into operation, for Welbeck Motors, a radiotelephone system which operates 200 radiotaxis initially, expanding to 800 by the end of the year.

The system with six-channel radiotelephone equipment is installed under the dashboard of each car and gives communication over a radius of 20 to 25 miles of Piccadilly Circus. A special control room has a con-veyor-belt message system designed to handle calls at the rate of one every two seconds.

## V.H.F. Unit for Lifeboat

THE President of the Dieppe Lifeboat Committee has been presented with a Pye International marine radiotelephone by the French National Railway,

This is the first French national lifeboat to be equipped with V.H.F. radio. It will work into the Dieppe railway radio station and the local public correspondence scheme. The lifeboat will thus be able to communicate with cross-channel steamers, with local fishing boats, which have the requisite equipment, and with the Boulogne Sur Mer Radio Station.

## Mobile Microwave System

THE United States Government has recently placed a contract with Marconi's Wireless Telegraph Co. Ltd., for the supply of a mobile microwave telephone and telegraph communication system to link many of the United States Air Force bases in the U.K. Marconi's and the Automatic Telephone and Electric Co. Ltd., who are also involved in the contract, are currently fulfilling this three-and-a-half million dollar agreement.

The Marconi radio equipment and the Automatic Telephone and Electric telephone carrier equipment is housed in semitrailer air-conditioned vehicles. Inside these large vehicles the transmitter, receiver and supervisory equipment are each contained in racks arranged for easy accessibility by the operators.

## Naval Plotting System

MOST of the Western nations were represented at a recent demonstration of a new naval automatic plotting systemdeveloped by E.M.I. Electronics Ltd.- given on board H.M.S. Rhyl, one of Britain's latest antisubmarine frigates. The system constitutes a big advance in plotting the positions of other craft in the vicinity, and is of particular use when hunting fast submarines. It is also of great value for coastal navigation.

In ships fitted with this system, the former method of manual plotting and the reporting of bearings and ranges over an intercom network has been replaced by E.M.I.'s electronic equipment, which produces much quicker and more accurate results.

## Radio Paging System

Phillip W. HOLLAND, M.P.
for Acton Constituency, visited the Ultra Electronics factory at Western Avenue, Acton, recently to inspect their radio paging system.

He became interested in this equipment when he saw it at the Crime Prevention Exhibition organised recently at Acton Town Hall, by the local police authorities.

Mr. Holland has a question tabled to the Secretary of State regarding the use of such equip-
ment by the policeman on the beat.

## Moscow Exhibition

THE Council of the Scientific
Instrument Manufacturers' Association of Great Britain states that its 19 members who took part in the combined stand at the recent British Industries Fair in Moscow are very well pleased with the results.

Firm contracts have been signed for 75 per cent. of their exhibits. Preliminary contracts for a considerable additional quantity of equipment to be exported immediately or in the near future have also been placed.

## Cable Across The Thames

A $£ 100,000$ contract from the South Eastern Electricity Board to reinforce the electricity distribution system between substations at West Weybridge in Surrey, and Laleham in Middlesex, has been awarded to Associated Electrical Industries Limited.

The project will involve laying four miles of 33 kV oil-filled cable and associated pilot cables. They will be taken across the River Thames in buried pipes. Two canal crossings are also involved.

The work will be carried out by AEI Construction (Cables and Lines) Division and will begin in July. The cable will be of AEI manufacture, from AEI Cable Division's Gravesend factory.


The interior of one of the vehicles which Marconi are supplying to the U.S. Air Force. The rack on the extreme left houses the transmitter, the middle rack the receiver, and the right-hand rack contoins the supervisory equipment.

## 20c/s TO 200ke/s ON FUNDAMENTALS WITH LOW DISTORTION AND STABLE OUTPUT LEVEL

$\mathcal{J}$

- HERE are four problems involved in the design of a variable low frequency signal generator: stability, waveform, frequency range, and output level characteristics. Of the three main audio oscillator circuits, the beatfrequency oscillator is suitable for general purpose applications, especially where a wide frequency range is required with a single tuning control. It has, however, 'sceveral serious drawbacks, including lack of stability, bad waveform at low frequencies. and R.F. in the output. These disadvantages may thll he overcome by careful design, but the resulting instrument is very large and very expensive.
The conventional $\mathrm{L}-\mathrm{C}$ oscillator is ideal for single frequency operation but requires an impossibly large tuning capacitor to produce the lowest ranges of a variable oscillator. In addition to this, it is difficult to obtain a frequency coverage greater than $3: 1$ in each band, which means greater 7 ranges are necessary to cover the audio spectrum of $20-20.000 \mathrm{c} / \mathrm{s}$.

The R-C oscillator, however, easily achieves a coverage of $10: 1$, reducing the number of ranges to three. It has the further advantage that the basic circuit satisfies the four design problenns set out above. The R-C circuit has therefore been chosen for this design.

## The R-C circuit

The basic circuit is shown in Fig. 1. It consists of an amplifier back-coupled through a wein bridge. The bridge is made up of the resistors


Fig. I. The basic circuit.

# WIDE-RANGE AUDIO OSCILLATOR 

By R.C. Englefield

$\mathrm{R} 1-\mathrm{R} 4$ and the capacitors $\mathrm{C} 1-\mathrm{C} 2$, with the innut and output of the amplificr connected to the diagonal points. By this means, both positive and negative fcedback are applied to the amplifier. The circuit oscillates at the balance frequency of the bridge. The negative feedback, through RIC1 and $\mathrm{R2C} 2$ determines the frequency, given by

$$
\mathrm{f}=\frac{1}{2 \pi \sqrt{(C 1 . C 2 . \mathrm{RI} . \mathrm{R} 2 .)}}
$$

The positive feedback, through R3 and R4, stabilises the level of the oscillation.

Since the input and output impedances of the amplifier are not directly connected across the arms of the frequency discriminating network, the amplifier characteristics have no cffect on the oscillator frequency.

In order to stabilise the amplitude of the oscillation with change of frequency, the attenuation of the negative feedback network must be kept constant, that is. the ratios R1/R2 and C1/C2 must he constant. The frequency of the oscillation may bc adjusted by alteration of cither pair of these components, and in this design, continuous frequency variation is achieved by varying the capacitances, different ranges being selected by variation of the resistances. A much higher degree


Fig. 2.-The complete practical circuit diagram.
of amplitude stability can be obtained by subatituting a metal filament lamp for the fesiolor R3. Thus when the amplitude of the oscillation falls, the positive teedback current decreases. so reducing the resistance of the lamp-this in lurn increases the positive fecdback and compensates for the original drop in amplitude.

## The practical circuit

A complete britge-stabilised circuit is shown in fig. 2. The first iwn stages ale based on the circuit of Fig. I, with the addition of the I5W 240 V lamp, 11, and range switching. The turing capacitor is a standard tworgang capacitur wilh trimmers, and the resistors RI-RY are luw tem-perature-coetticient types, thumgh ordinary tesis-
lose shomblat prove satistiactory in use. The range resisturs are best mounted ofl a ceramic switeh.

The output of the uscillatur is fed to the grid of a cathude follower, and the output from the unit is taken from a shorting-type microphone jatch. If the oscillator is reyuired to work into lotal impedances lower than 20k, it will be neeessuty 10 substitute a $25 \mu \mathrm{Fi}$ capacitor in place of Cil. il excessive luss of voltage is to be aroided at very low liequencies.

The power section is a conventional fiull-wave circuit, alal generous smoothing arrangements are ased. wilt the resull that the hum level of the out= pat is less than 0 - ?per cent of the maximum vutput voltage. Solsened wire should be ustad tor the wiring of the heaters and fanal lamp.


Rear view of the oscillator.

## Construction

Cunstltetion is simple and vrashtforward. The tumng capacilor is insulated from the chassis witl a sheen of Paxulint, and carries the puinter; slaw mothor turning is employed, a simple frictional drive being easy to make and pertectly satislachary. Ilar range switeh is mosunted benesth the tuningeontrol. and selects the folluwing ranges: 20 $10200 \mathrm{c} / \mathrm{s}(\mathrm{A}) 200$ 10 $2000 \mathrm{c} / \mathrm{s}(B) 2$ (1)20kc/s (C) 20 lu 200kic/s (D). the fange can be extended up to $1 \mathrm{Mc} / \mathrm{s}$ if nece revistor and a bsu!? resistor. mounted on a lifih pole of the range switch. A contimmonsly variable butput control. He vulput socket, the main switeh, and the indicator lampe complete the pathel items.

The scale is mate from Perspex sheat, mbunted in liomt of the pointer: Joce calibrations are dawn irs laclian ink onl the hach of the Perspex this type of suale being very duratule, shat has the advantage that eloors cañ tue toorrected isitho out Litue by removing the ala with



Hy bonnecling the scope 10 display Itissajous figutes, using $50 \mathrm{~L} / \mathrm{s}$ mains as standard, ranges A afted is can now be calibrated: points should be logged every $5 \mathrm{c} / \mathrm{s}$ in range $A$, and every $50 \mathrm{c} / \mathrm{s}$ in range $b$. To catibrate range $C$, set the internal timebase of the seope to $500 \mathrm{c} / \mathrm{s}$. again using the nitains as standard, and long points every $500 \mathrm{c} / \mathrm{s}$ throughout the range-it is recommended that the sweep frequency be checked directly after each catibration is taken. Lissajous rigures may agairt be used to calutrate range $D$ if a $100 \mathrm{ke} / \mathrm{s}$ quartz crystal standard or other ultra-stable oscilliator is avalable, thut if nol, points can be accurately obtained by beatiog the harmonits of the oscill


The tuning drive mechanism.
little af a thome keeping their eapacitues atmut equal, umil the 200e/s print is just at the end of the range. Ihis sets the figh ened of all four ranges.

Before commencing the calibration proper, remove the ferspex dial and construct a hogeing scale on the surface nearest la the ponter: His scale should have as large a ratiols ax the lengrh of the pointer will allow. A large promachor will simplify He devision of the cilcomiterence into 180 patis, het if this is mot available, llief a seale al divisions (say 256 ) may but constructed by contmonas bisectuan of a $300^{-}$- mis is hest drawn out on paper first. and then Handefored to the Perspets.

Fig.,5.-The front panel.

lator against the carrier waves of brosdcasting stations, especially in the long-wave band.

The scale may now be removed, and the dial calibrated directly in terma al frequency: lurther points may be obtained if necessary by graphical means; if any marks are nisplated, they can easily be corrected as explained.

The overall performance of this instrument is more than adequate firr ordinary testing: the output is inherently stable in frequency and level, and the hum level and distor= lion dre especially low.' These advantages more than offset the slight disadvantage that rangeswitching is necessary to cover the amtio spectrum.

## m

ANY amateurs have made up an R.F. generator and then found that the greatest difficulty, often insuperable, has been accurate calibration.
$\therefore$ Details will be given of methods of calibration -using broadcast stations as standards, but it is so simple to build a standard crystal oscillator that it is proposed to deal with the construction of a simple unit as part of this instrument.

The author housed the crystal check unit in the same case as the other chassis, but it could be kept separate and only plugged in or connected when required. Valves from the variable audio oscillator could be "borrowed" for this, but it is better to have two kept for use in the unit alone.


Fig. 27.-The crystal oscillator circuit.

BUILDING CHASSIS No. $4-$
THE CRYSTAL CHECK OSCILLATOR
(Continued from page 306 of the August issue)
By:E. V. King

## THE P.W. SIGNAL GENERATOR

Fig. 28.-The cirçuit of the multivibrator.
passing through the valves V6a and V6b-if the valves were operated for long periods without H.T., the cathodes' would become "poisoned" and the circuit would no longer operate.

Switch S10 is used to switch on or off the complete crystal check oscillator-Chassis 4 . When S 10 is on, the $100 \mathrm{kc} / \mathrm{s}$ oscillator will function.

The signals obtained are fed to V 7 b , a harmonicproducing amplifier (Fig. 30); the diode helps in this respect, too. If Si 11 is on, $10 \mathrm{kc} / \mathrm{s}$ signals and harmonics are thus fed to the output socket. When S 11 is off, the $100 \mathrm{kc} / \mathrm{s}$ signal from V6b goes through C26, C25 and C27 and is thus amplified in the same way. The $100 \mathrm{ke} / \mathrm{s}$ signal will thus be fed through to $V 7 \mathrm{~b}$ in addition to the $10 \mathrm{ke} / \mathrm{s}$ signal from the multivibrator (V6a and V72).

## Mounting the Components

There is plenty of room in the chassis and the layout is in no way critical. Fig. 29 shows the wiring for the $100 \mathrm{kc} / \mathrm{s}$ crystal oscillator.

Rotary toggle switches are used so that the fact that Chassis 4 is a check unit is apparent from the front, all other switches being of the lever toggle type. Likewise the output socket is at the back so that it cannot be used in error for the K.F. or audio socket.

Make sure the tag strip does not, short to earth and that the valveholders are correctly positioned as in Fig. 29.

## Inductance 15

This has an inductance (measured) of 11.5 mH , but is not critical.

A length of ferrite rod (see the Components List and Fig. 24) is covered with a layer of Sellotape and then 400 turns of 32s.w.g. enamelled copper wire are wound on approximately to the dimensions of Fig. 29 where distances "a" are $\frac{3}{8} \mathrm{in}$. and " $b$ " are $\frac{3}{4} \mathrm{in}$. Wind-

Note that tag 5 of the power unit is also used for the audio oscillator. but this does not matter as the two units will not be used at the same time.

Take one side of the crystal to earth and the other side to pin 2 of $V_{6}$. Join pin 2 to.pin 3 via R30 and connect C26 from pin 2 to the output socket (temporarily only). From pin: 3 take R31 to earth and C23 also to earth. Note. there are now three wires on pin 3.

## Testing the Crystal Oscillator

Connect it to the power unit. Put a milliammeter in the H.T. lead to tag 5. The current taken will be about $2 m$ if the unit is osecillating and SmA if it is not oscillating.

While the unit is connected, withdraw the crystal and the H.I. current should alter from the approximate value of $2 m a$ to about 8 mat. It
ditticulty is experienced and


Fig. 29.-The underchassis wiring of the crystal oscillator.
ings are pile-wound and must all be in the same direction. Baisa cement may be used to keep the windings in place. Check that the windings do not short to earth when mounted.

The rod may be mounted in any convenient manner. The author preferred to insulate it from the chassis. at the ends, but it worked quite salisfactorily uninsulated.

## Wiring the Crystal Oscillator (V6b)

The circuit is shown in Fig. 27 and the layout and wiring in Fig. 29. Proceed as follows: take pin ${ }^{1}$ and the centre spigot of each: valve to earth. Join pins 4 and 5 of each valve and wire them via Slo to a tag (2) of the tag: strip which is on top of the chassis. This tag strip is fixed in a similar position to the strips on the other three chassis-see Fig. 23 last month for example. Connect the unit to the power pack tags 1 (or 2) and 7. Plug in the valves and observe that all heaters light correctly.

Now take one end of the coil to pin 1 of $V 6$ and the , other end'to, a tag (5) of; the: tag strip.


Fig. 30.-The distorter amplifer or harmanic producer.
or hum, should be heard at the $200 \mathrm{kc} / \mathrm{s}$ mark (Light Programme) and also at $300 \mathrm{kc} / \mathrm{s}$ and $400 \mathrm{kc} / \mathrm{s}$. If the receiver will tune to $100 \mathrm{kc} / \mathrm{s}$, a stronger signal will be heard there. If the output is connected via a condenser to the aerial terminal, strong signals will be heard as the receiver is tuned through the $100 \mathrm{kc} / \mathrm{s}, 200 \mathrm{kc} / \mathrm{s}, 300 \mathrm{kc} / \mathrm{s}$ and $400 \mathrm{kc} / \mathrm{s}$ points.
1 The $100 \mathrm{kc} / \mathrm{s}$ point is the fundamental. It is now proposed to increase the number of strong harmonics so that the receiver may pick them up even on $25 \mathrm{Mc} / \mathrm{s}$. As it is now, the crystal oscillator will check receiver calibrations on the Long Wave band and part of the Medium Wave band.

## The Harmonic-producing Ampllifer

Thè circuit is shown in Fig. 30 and the layout and wiring in Fig. 31.

Proceed as follows: remove C26 from the output socket and solder it to pin 2 of the other valve (V7), and then wire pin 2 to earth via R37. From pin 3 of this valve take R36 and C28 to earth. To pin 1 connect the diode (do not shorten the leads too much or allow them to become hot), and wire the other end of the diode through R 35 to tag 5 on the tag strip (for H.T. from the power unit). Note that the red ( $t$ ) end of the diode is connected to pin 1 of V7b and the black (-) end to. R35. From pin 1 take C29 to the output socket. From the H.T. end of R35, wire R32 to pin 6 of V6.

## Testing the Amplifier

Connect 'phones to the output socket. Remove V6 from its holder, plug in, switch on, etc., and touch pin 2 of V7. A humming should be heard in the 'phones. Replace V6 and repeat the previously described tests with receivers. Calibration points will be found every $100 \mathrm{kc} / \mathrm{s}$ up. to at least

## COMPONENTS LIST

 FOR CRYSTAL CHECK UNITBaking Tin (as for the other chassis)

Resistors:

| R30 IM $\frac{1}{4} \mathrm{~W}$ | R35 | 47k $\frac{1}{4} W$ |
| :---: | :---: | :---: |
| R3I Ik $\frac{1}{4} \mathrm{~W}$ | R36 | $1 \mathrm{k} \frac{1}{4} \mathrm{~W}$ |
| R32 and R33 10k $\frac{1}{4} \mathrm{~W}$ | R37 | IM $\frac{1}{4}$ W |
| R34 5.6k or 6.8k ${ }^{\text {i }} \mathrm{W}$ | R37a | 470k $\frac{1}{4}$ W |

## Condensers:

C23 $0.1 \mu \mathrm{~F} 200 \mathrm{VW}$
C24 and C25 $0.005 \mu \mathrm{~F}$ ( 5000 pF ) Mica
(No other value is suitable-parallel wiring of say 1000 pF and 4000 pF condensers will give the required value)
C26 10 pF , mica or ceramic
C27 50pF, mica or ceramic
C28 $0.1 \mu F 200 \mathrm{VW}$
C29 $0.01 \mu \mathrm{~F} 350 \mathrm{VW}$
Valve V6 and V7 Mullard ECC8I
Switch SII Rotary toggle type
Switch SIO Rotary toggle type
Diode Any germanium or silicon diode (surplus)
Inductance LS About $10 z$ of 32s.w.g. enamelled copper wire and a $7_{16}^{5} \mathrm{in}$. dlameter ferrite rod $5 \frac{1}{4}$ or $5 \frac{1}{2} \mathrm{in}$. long is required

## Quartz Crystal Any 100ke/s crystal will suit,

 the prototype uses a rather large surplus type avallable from Henry's Radio. A holder is available alsoOutput Socket Normal co-axial type
4 or 4-way tag strip-two tags being earthed by fixing
Variable Resistor VR5 10k, any type will sult. A surplus locking ring is useful to prevent this control from moving once it is set
Nuts, bolts, grommets, etc


Fig. 31.-The wiring of the harmonic-producing amplifier.

## Building the loke/s Multivibrator

The valve heaters for this part of the circuit are already wired and tested. The circuit can be seen in Fig. 28 and the wiring, which has to be added to that already carried out (Figs. 29 and 31), can be seen in Fig. 32.

Unsolder C26 from its temporary connection on pin 2 of V7. Join pins 8 of both valves together. Connect pin 8 of V6 to one side of Sil and the other side of S 11 to earth. Take pin 8 of V6 temporarily direct to earth. Take pin 7 of V6 to one outer tag of VR5 and also through C25 (which must be of the correct value) to pin 6 of V7. Take pin 7 of V7 to earth via R34 (this must also be of the correct value). It will probably be easiest to earth pin 7
$25 \mathrm{Mc} / \mathrm{s}$ (i.e., even on the 16 m band which is considered adequate for this instrument). Note that most ordinary receivers are only approximately aligned to the scale on the short-wave bands.
of V7 to the centre spigot or to pin 9. Join pin 7 of V7 also to pin 6 of V6 via C24 (again, this must be of the correct value). From pin 6 of V6, join R32 (positioned upright) to a similar
resistor R33 (also upright) and the junction of these two (R32 and R33) is joined to R35 where it leads off to tag 5 (H.T.). The wire joining the three resistors is left " in the air". The other end of R33 is connected to pin 6 of V7.

## resting the Multivibrator

Connect up to the power pack with a milliammeter in the H.T. lead to tag 5. The current should be about 10 to 12 mA with both the amplifier and crystal oscillator working.

Connect to the aerial socket of a receiver via a small condenser (10pf) (and via C27) to pin 6 of V7. Loud and numerous "swishes" will be heard in many places on any waveband, the number of swishes being controlled partly by VR5 which must not be turned to the completely "off" (i.e., shorting) position.

## Synchronising the Multivibrator

Condenser C26 has already been connected to pin 2 of. V6. and is now not connected at the other end: solder this end to pin 7 of V6. The $100 \mathrm{kc} / \mathrm{s}$ signal will now synchrunise the multivibrator.
Connect up as before to a receiver set to long waves. Notice exactly where the $100 \mathrm{kc} / \mathrm{s}$ points were they will be the loudest "whistles" or "swishes") and verity that "swishes", previously nun-existent, are now heard in between the louke/s points. Verify that the number can be altered by means of VRS.

## Setting the Multivibrator on 10kc/s

Connect the unconnected end of C27 to pin 2 of V7. Disconnect the earth lead from pin 8 of Y6. The circuit is now as shown in Fig. 28.

It is vital that VR5 control is correctly set-giess-work is not good enough. Place the unit near to a selective receiver, and connect it to the aerial socket via a low value condenser (only if it : proves necessary). Disconnect the normal aerial. Switch to Leng waves and, tune in the Light Programme.. Switch S10 on and S11 oft. Note very carefully (a piece of gummed paper on the dial will help) the crystal calibrator point at $200 \mathrm{kc} / \mathrm{s}$ and tune in the next one at $300 \mathrm{kc} / \mathrm{s}$ and mark it accurately. Now switch on S11 so that the multi-vibrator functions. Count very carefully the "swishes" between the two points already marked. There should be nine (i.e. eleven counting the two main points). If the number is not correct, 'move VR5 slowly until you hear it lock to another frequency (in the receiver, or, if your hraring is good, from the vibrating valve electrodes. in the vibrator). Repeat this until you are sure. Near the actual $100 \mathrm{k} / \mathrm{s}$ points, the hiss or whiscle may be so loud that it spreads over some of the dial and masks what might be another "swish" or "whistle".


Fig. 32.-The wiring of the multivibrator.

The remedy is to move the generator farther away from the receiver so that the "swishes" are not so loud.

Finally, when VR5 has been correctly set, there should be nine smaller "swishes" between the two larger "swishes". If the receiver used is not sufficiently selective, then only seven smaller "swishes" may be heard. The two which are nearest the two larger "swishes" may be hidden by them. However, it should be obvious that two are beirig hidden because the spacing between the seven smaller "swishes" will be regular, while the spacing between the final smaller "swishes" and the two larger "swishes" will be noticeably-greater: it will be obvious that there is space for the two "swishes" which are masked.

When VR5 is correctly set it is sealed; sealing wax is suitable.

## Checking Receiver Dials

The receiver dial may now be checked; approximately, using the $100 \mathrm{kc} / \mathrm{s}$ points (these are, however, exactly accurate), and then in detail with the loke/s points throughout the whole range of most ordinary receivers. The beginner, should experiment in this direction to gain experience.
(To be continued)
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## Those Portables

SOME time ago I mentioned the annoyance which was being caused by the transistor type of portable and suggested that perhaps legislation would have to be introduced to limit the use of this latest weapon. Since then. I was very relreved to see that a number of authorities have taken steps to preserve the rights of the majority of us who prefer quiet and peace. At Lord's cricket ground the use of the portable has been banned, as well as on the sun-deck of the pier at Southend. At least one other seaside resort has a bye-law preventing the playing of radio in the streets after a certain hour and one visitor who was walking along with his radio playing was fined.

These little sets are wonderful, if properly used, but now one sees them, in my view, being abused. On an underground train recently, one gentleman was tinkering with one and when he got out at my local station (which was above ground) he slipped it in a small cloth shopping bag, still playing, and his taste in music was inflicted on all the passen. gers, and presumably also on the pedestrians in every street through which he passed. The recent spate of cricket and tennis relays has brought untold misery to many and I do feel very strongly about these receivers, and think that the type with a plug-in earpiece deserves wider use as then others do not have to listen. It is surely a very selfish attitude to use these loudspeaker portables in such a way that others can hear them. and I feel sure that their continued use to the annoyance of others will cause so many complaints that -ventually they will be banned from public places. or at least a limit will be put on the volume of sound which may be used. This, of course, brings in the problem of our old friend the decibel. When a standard such as this is adopted, some measuring device must be employed and users of road breaking apparatus and others who may fall foul of the law may plead that the volume was not excessive whilst those who object contradict this, so without a reliable yardstick, legal actions will not be possible. Perhaps the Noise Abatement people will produce some instrument which may be legalised as a measurer of noise, and receive police sanction, so that those who wish to complain will be able to do so with every chance of being upheld in their complaint.

## Speed Detectors

The talk of noise measurers, brings to mind the speeding car detector, and it seems that our lives
are being rather filled with modern scientific apparatus to the exclusion of our own senses and abilities. The modern calculating machine, the instant photo copier and similar devices are now commonplace, and whilst there is little doubt that they do improve efficiency by speeding up the result, the human endeavour is ousted and 1 wonder whether this is all to the good. We shall not need the clever mathematician in future years, as the calculator will take his place, and the efficient typist will be pushed out by the electrical typewriter, complete with automatic gadgets, whilst the skilled photographer who is now called upon to make accurate copies of documents, etc., will be replaced by the "copier".

## Amateur Satisfaction

My postbag contains letters on a multitude of subjects, but it is surprising how many deal with the radio of earlier days. As one who saw radio commence its entry into the world, I do agree with a correspondent who wonders whether the modern amateur gets as much satisfaction from modern equipment. He sends me a neat pile of leaflets which he found amongst some gear at home and sends them for my inspection. He says

From my point of view, the list of the Northampton Plating Co. was of particular interest, as I remember that it was one of those sources of components at a price low enough for me to afford, being a schoolboy at that time. Their S.L.F. variable condensers at 3 s 11d were well made and smooth in use, one of them was surviving until some ten years ago when it was passed on to some local Scouts with other items for further use. It was still quite free of wear or shake. Incidentally S.L.F. type single condensers do not seem to be very easily obtainable nowadays, which is a plty in view of the number of circuits appearing for small receivers and alignment oscillators in amateur publications. Most condensers offered at "popular" prices seem to be either S.L.C. or law types, and twin gang at that.
"The dull emitter valves advertised by the Plating Co. were also quite good for their time, although the two 1 purchased had a greater tendency to microphony than the Mullard and Cossol products.
"Still, one sometimes wonders whether the present day schoolboy, with a range of components available at relatively low price, gets quite the same gratification as my generation did at res ceiving Schenectady on a crystal and catswhisker."

Yes, I wonder whether the same degree of satisfaction is experienced, especially by those who received tuition in radio at school and therefore regard it not so much as a mystery but as an accepted thing-like the essential services to be found in the home?

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## AN

ALLL TRANSISTOR PORTABLE RECEIVER

Designed by D. R. Bowman

7HE design of the receiver described here is admittedly a by-product in the development of a fully-transistorised television receiver. Before the latter can be presented, a number of serious problems remain to the solved: but the radiofirequency and intermediate-frequiency stayes proved very amenable to normal transistur design procedure, and it was therefore decided to make the fent necessary moditications ou uffer a fully transistorised VIft receiver. Work with this will atfiord valuable experience to the constructur, as it did to the writer.
It was reatised from the butset that the production of ar receiver of the two-hanal miniature type would result int the ahaw twoustat "tirnty" reception, and this would do far lens than justice hoth to the excellent and imterference-tier VIFF transmissions and to the present-day high quality triansistors. Provision has thereture been made for class AB audio output up to IW, and for reasomable reproduction of bass trequencies, a miftimum
size of cabinet had to be specified. The receiver is certainly quite portable, as it weighs only a few pounds, but is not in the vest-pocket class.

Sensitivity is of the same order as the usual hind of domestic F.M. mains-operated receiver. An aerial signal of $5 \mu \vee$ gives a power output of 50 mW with the prototype receiver. the normal spread of transistor parameters may cause this to vary appreciably-perhaps from 3 to $20 \mu \mathrm{~V}$-but this will be found quite adequate in practice. Signals of 200 mV can be handled, but where such a strong signal is obtainable, a very sketchy aerial can be used, and doubtless will be used, by many constructors. An vutside uerial, 20 ft high, consisting of a plain dipole, will ensure reception at distances of 100 miles or mure from thee transmiter.

In the prototype receiver, prinied-circuit con* struction has been employed. The circuit will be fund readily adaptable however fur those who wish to use a normal chassis, but the use of copperclad tammate is recommended for its ease of worhing and its goud electrical properties-ino difticulty will be experienced in the etching process.

Fig. I shows the cireuit diagram. This, it will be seen, denutes a "line-up" of nine ransisturs; an K.t amplifying stage, self-oscillating frequencychanger, three 1.1. slages, ratio detector, audio preo amplitier, driver and push-pull ourput stages.

## The R.F. Stage

The R.F. stage consists of a Mullard OC171, operating in the grounded-base configuration. The real part of the emitter input impedance under the conditions of the circuit, is about 1202. The aerial is therefore connected direct to the emitter through the usual coaxial cable of nominal impedance $80 \Omega$ This provides a very reasonable match. If, how ever, a length of wire is used for portable operation, best results are obtained by plugging a 42 in . length into the centre connection of the coaxial input socket. It will be remembered that a single 33 in . length will present an impedance of about $40 \Omega$ to the emitter, and the extra length is required for a better match.

The input is of course not aperiodic, as might be supposed: the input tuned circuit is in fact the aerial. As long as its Q is about 12 or less, the sensitivity drop at the edges of the VHF band will be less than 6 dB . The aerial should therefore ideally consist of a dipole using copper or aluminium tubing of $\frac{1}{2} \mathrm{in}$. diameter. However, the writer uses a picture-rail aerial consisting of two 33 in . lengths of $14 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. copper wire, with good reception of the "local" station ( 56 miles away).
The R.F. OC171 operates at a collector current of about 1.4 mA . A higher current gives negligible increase of gain and a noticeable rise in noise-level. A capacitor connected between emitter and base, of value 20 pF , may require some comment. The purpose of this is to bring the collector feed-back current more into phase with the aerial current, and results in a measure of positive feedback, increasing R.F. gain by 3 to 4 dB . It may be pos-

The collector circuit is tuned partly by a fixed capacitance of negative temperature coefficient and partly by a variable capacitor in series with another capacitor of negative temperature coefficient. The collector transtormer is so designed that the secondary winding matches the input impedance of the frequency-changer transistor.

## The Frequency-Changer

The frequency-changer, another OC171, is also operated in the grounded-base configuration, current from the R.F. stage being introduced by including the secondary of the inter-stage transformer in the emitter circuit. As this is a selfoscillating frequency-changer, and feedback from the collector is arranged by capacitive coupling, it is necessary to include a resistor of about $100 \Omega$ in the emitter circuit also. This causes some reduction in the signal injected, but the effect is small. The oscillating circuit is tuned partly by a pre-set capacitor, and partly by a variable capacitor ganged with the R.F. inter-stage tuning capacitor. Both are in series with a fixed capactor of negative temperature coefficient which also does duty as the fixed capacitor of the I.F. transformer in the collector circuit.
The use of negative temperature coefficient capacitors in the R.F. and F.C. circuits is not-as with valve operated equipment-to correct for warming-up difift. In the transistor receiver their function is to correct for changes in ambient temperature. and conditions are much less critical. If the user does not object to re-tuning as the room temperature changes, silver-mica capacitors can be used instead.


Fig. Ia.-The input and I.F. stages of the circuit.
sible to increase this capacitor to a value of 25 pF , but in the prototype receiver such an increase caused the R.F. stage to oscillate when the collector circuit was tuned to $95 \mathrm{Mc} / \mathrm{s}$ or higher frequency. By the same token, with certain transistors, it may be necessary to reduce its value to 15 pF or even 10 pF .

## The Intermediate-Frequency Amplifier

It was realised that many constructors would be unwilling to undertake the construction of a receiver in which the setting-up and alignment would require a valve voltmeter or oscilloscope. In this circuit, adjustable coupling has been avoided wherever possible, and the I.F. transformers bave
been arranged for fixed capacitive coupling between primary and secondary, with non-critical spacing between. windings. The only variable coupling is in the collector circuit of the last I.F. stage, where a simple drill enables the precise conditions to be reached, using a meter, or "nearenough" conditions by ear.

Mullard transistors, type OC170, are employed in the I.F. stages. Two pairs of coupled circuits are used between the first and second I.F. stages and between the second and third. Coupling is "critical", and the dynamic impedance of primary and secondary have been calculated to give the bandwidth needed. For exact bandwidth the transformers should be tuned by 160 pF capacitors; but 180 pF capacitors are here specified because the inevitable slight inaccuracies in practical alignment, result in a very slightly "stagger-tuned " amplifier, which still has the correct bandwidth.

The last I.F. stage, which drives the ratio detector, requires tighter coupling than critical, and coupling of approximately 1.5 critical is used. The intermediate frequency is $10.7 \mathrm{Mc} / \mathrm{s}$ nominally, though the constructor may prefer to vary this a little one way or the other to avoid a powerful local short-wave transmitter.

## The Ratlo Detector

The use of a ratio detector was decided upon because of its simplicity and its ability to deal with small signals effectively. Because of this latter consideration, AGC has not been provided and its absence has not been acutely felt. The

## Neutralisation

As the transistor may be regarded for some purposes as a triode with the grid operated in the positive region, it will be understood that at the intermediate frequency concerned, $10.7 \mathrm{Mc} / \mathrm{s}$, neutralisation of the collector-base feedback will be needed. By suitable design of the I.F. transformers, fixed non-critical capacitive neutralisation can be achieved. In fact, this circuit possesses very slight over-neutralisation, and with certain transistors the 5 pF neutralising capacitors may need reduction to $3 \cdot 3 \mathrm{pF}$ if I.F. instability is experienced. If desired, the addition of small resistors (about $47 \Omega$ ) in series with the neutralising capacitors will improve stability and gain. This process, known as unilateralisation, enables wanted and unwanted feedback to cancel each other out exactly in magnitude and phase. However, this is a refinement found not to be necessary in practice,although the purist may prefer to include it in his receiver.

## The Audio Amplifier

No claim is made that the audio amplifier is original. In fact it is almost identical with the Mullard .1W amplifier, described in "Reference Manual of Transistor Circuits" (first edition, 1960). There seemed to be no point in gilding the lily. However, with the transformers specified, crossover distortion at low signal levels was more noticeable than might be desired, and a slight increase in output quiescent current has been arranged to overcome this.


Fig. Ib. -The detector and audio stages of the circuit.
ratio detector is of conventional design. Diodes supplied in a matched pair (OA79) avoid the need for adjustable resistances in series. Amplitude limiting is improved by the stabilisation of only about 90 per cent of the developed voltage, and this is arrived at by the inclusion of small fixed resistors in series with the diodes. The normal de-emphasis (in this case, $45 \mu \mathrm{sec}$ ) is provided.

Very thorough decoupling has been found essential in the audio section of the receiver. This is hardly surprising, since approximately 70 dB audio gain is obtained. Layout however is not critical. The inevitable increase in battery resistance during discharge causes no instability. and as much as $300 \Omega$ has been put in series with the battery without causing trouble.
(To be continued)

# A Pre-wired Valveholder Extension 

## A SPACE SAVING MODIFICATION

By D. J. Gill

On
READER'S letter in the Fehruaty issue, on the suhiect "Extension of Valveholders," rempts me to submil this simple and effective solution, which I adopted some years agn.

This simple method not only chininates the cumbersome tagboaid. thus enabling the amateur in reduce the nverall sise of the mit under Construction. or take-advantage of the space saved to accommodate hetter his other components, huf it has the added advantage of allowing earh valveholder and its associated components th he wired and checked as a separate sish-unit before the valveholders are mounted on the chassis. as will be seen. Added to this is the advantage al being ahle to wite all valveholders while they are on.a. clear bench with every component readly nccessinle, as, opposed to having to probe into


Fig. I. (left)—The dimensions of the Paxolin disc.
Fig. 2. (right)-Two suitable extension pillars.
hidden sornets with a soldering iron, sometimes at the expense of wiring, or other components, only to finish up with a doubtful comnection.

Reference to the diagrams will show that the only fools required are a souple of taps and dies, hut even these are not ahsolutely necessary. Fig. 1 shows' the type of dise used by the writer, but the shape is optional, and end cheeks from discarded coil formers, or any other thin Paxolin or Bakelite will serve equally well. The dimensions for Fig. 1 can alsn he left to the reader, though those shown have been found to be more than adequate to accommodate all the components associated with any valve. When the size and shape has been decided, they can be drilled in batches of about six at a time as shown.

The extension pillar, Fig. 2, can be made from brass, dural, or one of the many thin rods or tuhing easily ohtained. For those whose kit does not include taps or dies, this pillar can be made ryually well from a OB.A. I lin. long screw, in the case of a mine pin valveholder. Fir a seven pin valveholder, the procedure is the same but the screw is 2B.A.


Fig. 3.-The simple construction of the extension.
The head of the sctew is removed and a GB.A. nt AB,A. nut soldered on one end of the screw. The thread is then hled off ahour a fing at the ohber end. This will leave the screw a tight pushfit into the centre of the valveholder.

Fig. 3 shows the unit ieady for assembly, and needa no cxplanation further than to say that the


Fig. 4.- The valveholder extension in use.
top dise can he left off until all the components have heen soldered to the valveholder pins.

When all components have been soldered to the valveholders. the free ends are slipped through the eyelets and the outgoing leads with them, the disc having been screwed in position. The reason for putting the outgoing wires in at this stage is that it is easier to put two or three wires if necessary through an eyelet before soldering, although there is nothing against making small loops in the component wires after they have been passed through the eyelets, and taking the wiring from these.

# Record-player and Radiogram Faults, <br> CAUSES AND CURES OF NOISE DISTORTION, WRONG SPEED, ETC. 

RADIOGRAMS can suffer from a number of troubles in addition to the usual radio receiver faults with symptoms peculiar to themselves. These defects are generally either pick-up faults or turntable faults.

## Piek-up Faults-Hum

Hum is more pronounced on radiograms, when present because of the larger speakers fitted and the extra attention given to the audio circuit by the designer to increase bass response. The larger baffle area for the speaker also accentuates this. Never attempt to clear hum troubles on a gram chassis without the accompanying speaker connected and fitted in its' cabinet. If this is not possible, then make sure that the bench speaker used is well baffled, because a low-level hum which is tolerable on a small speaker turns out to be overpowering when the chassis is fitted back in its cabinet.

Hum that occurs when the gram is switched to records, and is unaffected by the volume control, is residual and is introduced in the circuit after the volume control. Deteriorating smoothing capacitors and partial heater to cathode leaks in following audio valves are the most likely reasons.

Variation in hum level when the pick-up or motor board are handled denotes incorrect earthing. Check the earth connections under the board but do not add extra earths to chassis in the case of A.C./D.C. grams or the motor board might become live to mains. Some A.C./D.C. grams give less hum with the mains plug a certain way round in the supply socket (taking the chassis to neutral).

## Pick-up Faults (Fig. I)

Distortion on gram only, can be caused by the stylus or by the pick-up crystal itself. The stylus can be checked on rotating cartridges by reversing to the other stylus; if this gives undistorted results (on the correct speed record), then the other stylus is proved defective. In any case, it pays to replace worn styli before evidence of wear is audible. See that the stylus is properly located on the crystal transmission pad. Too often, through mishandling and heavy set-down on auto-changers, the stylus becomes dislodged and is found jammed between the cartridge side and pad (see Fig. 1). Debris pic...ed up as dust should be cleaned periodically from the stylus.

## Cracked Crystals

Good reproduction on radio but weak distorted


Fig. I.-Typical faults which may develop in the pick-up arm.

Hum which reduces with the volume control setting on "gram" only nearly always originates from the pick-up. This can be checked by unplugging the pick-up leads from the chassis when the hum drops to a tolerable level. Check that the pick-up leads are not reversed-the screen should be earthed or connected to the chassis-this is a common trouble as many grams are not provided with polarised sockets and the plugs can be fitted the wrong way round by a novice without affecting the sound but causing hum.
sound from records is often due to a cracked crystal. This can be checked by unplugging the piek-up plugs from the chassis and fitting a crystal microphone or a substitute crystal pick-up from a portable record player. Good results from any of these shows the original crystal to be at fault. Crystal cartridges are fragile and can be fractured by careless handling-it is not generally known however that they are prone to failure with high temperature and portable mains radiograms with upper vents discharging heat into the space under


Fig. 2 (above). - The styli may become worn if the balonce of the arm is malodjusted.

Fig. 3 (below), -Howling is caused by mechanical vibration being picked up and circulated by the ocoustic feedbock loop.


## Microphony

Howling which commences when the volume is increased above a certain minimum (similar to amplifier acoustic fesdash in a hall is duc 10 the cabinct condacting vibations from the loudspeaker to the turntable and record where they are piched up by the stylus. amplified. and sent round the feedhack lnop again (see Fig. 3). This is nsually a fault with bome constructed equipment where the motor hoard is unsprung of badly fitted. A rigid cabinet with haffle bracing reduces mechanical vibation and raises volume at which the threshold of howl is reached. On manufacturers equipment this trouble is more often due to tight, unslackened. turntahle transit screws.

## Turntable Speed

The usual fault causing wow, slurring, and slowing of the record is not the furntable at all but the record slipping and skidding on its label with the one in contact with it (1.ig. 4). This loss of friction can catuse a wabling noticed on susbained high notes or severe slowing and at fincs stopping of the tor record while the underside pile continue to rotate normally. The upper record when examined will he found to he sallect-shaped or la possess a suall eruption on the latel lifting the opper record clear of the main friction plane. A greasy of polished rim on the jockey wheel can canse the same effects hutt this time the forntable will he folotd in te furning slowly (Fig. 5). Check the pressure sping pulting the jockey wheel on (1) the pullev for reduced tensinn.

Permianent steady slow speed is mosi usually calsed hy engagemient of the wrang pulley. The pulley hush can slin down the monor spindle if the grub screws loosen accidentally and with litfle fy ywheel momentum, wow or flitter may be present to iron if out.

## (Continued on puge 407)

the lid "cook" rick-ups if the gram is operated for an hour or two with the lid down.

One of the commonest troubles with turnover pick-up cartridges is the breaking off of the fine wire leads at the termibals on the cartridge due to the constant flexing with retation. Do not solder dirert to :the ferminals thit remove the clips first and solder to them Oinerwise the beat mav damage the crystal. This trouble is sometimes intermittent if the broken lead niakes and breaks with vibration.

Trishfficient pick-up weight can also cause cutting out of the somind owing in the stylus tip being held clear of the bottom of the gronve. An adjustable lowering stop at the hase of the pick-up can allow the stylus to drop lower, or if the arm is almost weightess, the comnter balance spring also situated under the arm at the hase can he slackened off one hole. Too much weight however causes excessive wear of records and styli (see Fig. 2).


Fig. 4 (above).-Slow speed, slurring and wow, is usually caused by a warped record. Fig. 5 (below).-Some of the causes of permonent slow speed.
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# An Intercom Amplifier 

By V. E. Holley

## THIS AMPLIFIER WILL

 give reliable service AS A GENERAL COMMUNICATIONS UNIT.$J$HIS amplifier is designed for the reproduction of speech and it will give long and reliable service as an intercom. baby alarm, ctc. Power consumption and heat generation are both very low and it is therefore especially suitable for continuous duty over long periods. The prototype has been in service for more than 3000 hours without valve replacement or deterioration in performance.

 -



Fig. 2.-Details of the underchassis wiring.

## Output Stage

This stage uses a high gain pentode: $6 A C 7$, in the circuit given in. Fig. 1. Though not properly an output valve, the 6AC7 performs that function very well, with an output approaching one watt for a signal of about $2 V$ peak on the grid. The output transformer in the anode circuit should have a ratio of 80 or $90: 1$ for a $3 \Omega$ loudspeaker. Both cathode and screen resistors are bypassed for maximum gain.

## Vottoge Amplifier

The signal tequired hy the output stage is derived from the anode circuit of $\mathrm{V1}$, a resistance


Fig. 3.-The "talk-listen" switch circuit.
coupled A.F. pentode (6]7). The coupling capacitor, C 3 is given a value of $0.005 \mu \mathrm{~F}$ to introduce some attenuation of the lower audio frequencles which is desirable for clean crisp speech. With anode and screen loads of 470 k and 2.7 M respectively, and a eathode bias resistor of 2.2 k the gain of the stage is 180 and the upper Trequency limit is around $5000 \mathrm{c} / \mathrm{s}$ : In the grid circuit is a microphone transformer, 1!, having a step-up ratio of 1:100. so. that the external lines can he at low impedance and therefore less susceptible to hum pickiup. The screen of V 1 is decoupled to the junction. of 'R4 and R5. This avoids degeneration and loss of gain which, RS would introduce if the decoupling were to chassis.

## Negative Feedback

The overall gain of the two stages is more than is required and is reduced by negative feedback taken, from the output transformer secondary to the junction of R4 and R5 in the cathode circuit of VI. The feedback, and, therefore, the gain, can be adjusted as necessary by varying the value of R10. in the range 220 to 10008 and if the gain cannot be reduced to the desired working level. a further reduction can be effected by rémoving C5
With equipment of 'this sort it is found that if the gain is fixed-and constant. the user rapidly becomes accustomed to it so that the results are always consistent; it might be compared in this respect with a telephone. No external gain control is therefore provided.

## Power Supply

The H.T. required is 14 mA at $250-300 \mathrm{~V}$ and the total heater current including the indicator lamp, is is a little less than 1A. A miniature mains transformer of the instrument of television convertor type is used, together with a half-wave, contactcooled, rectifier. The power consumption from 240 V mains is less than 15 W .
The resistor R +1 , and the capacitors C6 and C7, provide the main smoothing, supplemented by R6 and C4 for the supply to V 1 . It must be mentioned that the elimination of hum requires rather special attention. A background hum which would pass unnoticed in a domestic radio receiver can be prominent and objectionable in apparatus where the sound output is normally zero, especially in quiet surroundings. In this amplifier, adequate smoothing and negative feedback give a silent background.

## Construction

Fig. 4 gives a plan of the chassis on which the prototype was built. It is not essential to adhere to this, but, whatever layout is adopted, it must be arranged that there is a distance of not less than about 6 in . between the mains and input transformers and that the magnetic axes of the n ains and output transformers are at right angles. The output transformer also. must not be-less than about 4in. from the input transformer and the latter should not be bolted to the chassis until after it has been wired up and orientated to the position of minimum hum pick-up-this position is usually quite critical. A wiring diagram is


Fig. 5.-The construction of the "tolk-listen" switch.
ner in two places and the speaker is supported at the rear by a light aluminium bracket.

## Components

All the capacitors except C2 and C5 should he 350 VW as they have to withstand the full H.T. voltage from the rectifier while the valve cathodes are warming up. The resistors can all be $\frac{1}{2} \mathrm{~W}$ except R11, which should be 1W. R9, which has a non-standard value of $160 \Omega$ is easily found by


Fig. 4.-The drilling dimensions for the chassis.
given in Fig. 2. It is convenient to omit the connection between the output transformer and RiO until the test stage is reached.

The speaker in the prototype is a 7 in . $\times 4 \mathrm{in}$. elliptical unit mounted on a piece of hardboard 10 in . $x 6 \frac{1}{2}$ in. in which a suitable aperture has been cut. The board is bolted to the chassis run-
measurement from a batch of $150 \Omega$ components of usual tolerance. V1 may be a 6 J 7 , $6 \mathrm{K7} 7$. 6 SJ 7 , 6SK7, or any direct equivalent, without alteration of component values and with very little difference in performance, though the use of a valve with a top grid cap and a metal body is recommended.
(Continued on page 414)

# Ferrite Hod Aerial "wame ( 

## CONVERTING A TRF SET TO A RECEIVER USING A FERRITE AERIAL AND ONE STAGE OF TUNING

$\square$HIS article describes a method of converting a simple TRF teceiver using a tuned R.F. stage with air-cored coils, to a receiver with a fertite rod aerial and only one stage of tuning.

The conversion was carricd oult on a miniature three valve mains receiver during the course of a complete overhaul, the valve line-1sp being three Z.77's used as the R.F. stage, the A.F.: amplifiet and the output valves (plus a crysial diode defector). The main advantage of the conversion was the elimination of the inconvenience of having to
ventional and the valve drew an anode current of thout 9 n A . The values of inductance against the secondaries of the coils refer to the medium-wave sections of the windings ouly.

## Choice of New Circuit

On looking at Fig. I it might be thought that the simplest thing to do would he to wind a coil on a ferrite rod, with inductances the same as for the ait-cored secondaries, and wire it in the grid circuit; leaving the rest of the circuit unaltered. However. such a solution is neither necessary nur practical. It is not necessary sinice with two similar cascaded funed circuits: as are present in Fig. I, the effective $Q$ of the rombination is approximately dounle that of the $O$ of ane circuit considered separalely; hence with a $Q$ of 60 for each circuit the overall $O$ is aboul 120. With a ferrite rod aerial it is yinite easy to construct a coil with a medium wave $Q$ of several hundreds - in fact the acrial constructed by the writer had a measured $O$ of 300 at $1.5 \mathrm{Mc} / \mathrm{s}$. Hence, it is

Fig. I.-The R.F. stoge of the receiver.
provide even a short length of aerial wire when the set was transported from room to room. A further, although not so obvious. advantage was the elimination of the anode tuning of the R.F. stage. This meant a very simple alignment procedure. 'This second advantage was gained because of the extremely high unloaded " $Q$ " of the Ferrite rod aerial.

The circuit of the R.F. stage of the receiver was as shown in Fig. I. The circuit was quite con-

Fig. 2.-The circuit of Fig. I with the alterotions necessary for the use of a ferrite aerial; the resistor previously used for anode decoubling is now employed as the anode load.

unnecessary to increase the effective $Q$ of the R.F. stage by employing anode tuning.

It is not practical to tune the anode of the R.F. stage since, with most double-tuned TRF receivers, the anode and grid coils are not separately screened, but merely screened from one another by being placed one above and one below the chassis. With a high $Q$ grid coil it is extremely difficult, without special precautions in layout and screening, to ensure that positive, uncontrolled, feedback does not take place. Since this was a conversion, the layout had been already fixed and the slight increase in selectivity which would be obtained did not justify extensive modifications to the existing layout.

The circuit so far described is an R.F. stage with grid tuning only. A further simplification made in the writer's set was to restrict the tuning to medium-waves only. This was decided since the set was mainly intended for domestic listening and the only station normally required on long waves was the BBC Light Programme; Since this is easily obtainable, in the writer's location, on medium waves, the inclusion of long waves was not considered justified. In order to use all the control positions of the set, the hole previously used by the wave-change switch was used for the "on-off" switch and a volume control without a switch was employed.


Fig. 3.-A suitable method of mounting the aerial.
A difficulty which arises on medium waves (without a tuned anode circuit), particularly at the high frequency end of the band, is due to the stray capacities appearing across the anode load of the R.F. stage. The effect of such capacities is to shunt the anode load and so reduce the gain of the stage. This is unfortunate, particularly if it is required to receive Radio Luxembourg since this station is fairly elusive anyway. However, the difficulty can be easily overcome by employing the anode coil (previously tuned by the variable condenser) in series with a resistive anode load, and peaking this coil on about $1.5 \mathrm{Mc} / \mathrm{s}$ by means of its trimmer. Hence, instead of the gain falling at the H.F. end of the band, it can be made to rise and then fall away sharply. No difficulty was experienced with unwanted feedback since the
fixed anode tuned circuit is quite isolated from the variable grid tuned circuit, and no long grid and anode leads need to be adjacent. With a coil inductance of $160 \mu \mathrm{H}$, to tune to $1.5 \mathrm{Mc} / \mathrm{s}$, the preset tuning condenser must have a value of at least 60 pF ; this assumes a stray capacity of 10 pF . In order to be sure of satisfying this condition a fixed condenser of 20 pF was added in parallel with the 60 pF trimmer already available. The circuit arrived at from the above considerations is shown at Fig. 2. It can be seen that the resistor previously used for anode decoupling is now employed as the anode load.
The aerial coil was wound on a piece of 3 in. $x$ $6 \frac{1}{2}$ in. ferrite rod, 52 turns being required for an inductance of $160 \mu \mathrm{H}$. In order to achieve the high Q, necessary the coil was wound with Litz wire. A suggested method of mounting the aerial is indicated by Fig. 3.

Since the coil was wound with Litz wire, care had to be taken to be certain that each strand of wire was soldered on to the tags. An extremely satisfactory way of cleaning the ends of the Litz wire is given as follows. A small tin lid is filled with "meths" and the "meths" is ignited: the end of the Litz wire is first placed in the blue flame. The wire is held here for a few seconds allowing the cotton to burn off, and the copper to become red hot. The wire is then plunged into the " meths" and then immediately withdrawn over the side of the tin lid. If the method has been carried out correctly, each strand of the Litz wire will be perfectly clean and show the characteristic pink colour of unoxidised copper. As soon as the wire has been withdrawn from the " meths" it should be tinned.

The receiver should first be tuned to a station at the L.F. end of the dial, and the aerial trimmer and tuning condenser adjusted until the station coincides with its marked position on the dial. The set is then tuned to the required H.F. station at which the anode tuned circuit is to be peaked (e.g. Radio Luxembourg) and the anode trimmer adjusted for maximum volume.

## Record-player and Radiogram Faults

## (Continued from page 400)

Generally quiet passages or sustained notes show up this effect which is heard as a flutter or gnawing sound. It may be due to a fault in the rotor causing uneven pult, or to a flat on the drive pulley.

## Groove Jumping

Microgroove records in good condition that exhibit this trouble at any random place across the record point to excessive lateral stiffness of the pick-up arm. See that the pick-up screened wire is flexing freely without resistance and that the central bearing holding the arm is lightly lubricated.

Groove jumping that takes place always near the end of the record is often caused by the trip mechanism becoming stiff. When examining this small, lightly riveted pawl, under the turntable do not oil it. Clean it and exercise it with a few twists with the fingers to clear any chemical corrosion that can increase friction but to oil it will only attract the dirt and dust and cause the same trouble later.


# A <br> WOB 

 for
## R

ECEIVERS for the GBC's FMI. service have always presented difficulties of alignment for the amateur. The usual methods emplny an A.M. signal genciator and a sensitive voltmeter. In such a method, the generator is connected hetween the chassis and the input of the receiver's $1, \mathrm{I}^{\text {F }}$. amplifier, and the meter hetween the chassis and the grid at the limiter valve. The cores of the $1 . f^{2}$. transformers are adiusted for maximm readings on the meter. The meter is then moved to the discriminator stage and the alignment completed.
various frequencies. and the output voltages are moted for each. A curve may then he drawn showing ontput voltage against frequency: a similar procedure may he adopted with the discriminalor. However, should the carves have the wrong shape, then when steps have been taken to correct them, another set of curves must he drawn. 'This alignment procedure is obviously not ane to he recommended for the amaleur.

What is needed is a visual display of the varinus characteristics without the necessity of drawing graphs. The instrument for this display is the uscilloscope-as in the graplos which may be drawn, the $y$-axis is used to represent frequency and the $y$-axis to represent the output voltages. In an I.F. amplifier may be plotted on graph paper hy using an accurately calihrated signal generator ill comiunction with a meter: Unmodnlated signals are meter. Unmodnated signats are fed into the I.I. amplifiet at


Fig. I.-The circuit of the wobbulator. Note: the value of the resistor R4 is 6.8 k .

Whilst this procedure is satisfactory for approximate al!gnment, it is not always good enough, as I.F. amplifiers in F.M. receivers need to have a certain handwidth if the signal from the transmifter is $t 0$ retain its high fidelity. The discriminator especially needs to he aligned accurately so that ins characteristic is linear if distortion is not to be heard in the ouput of the recciver.
use. a frequency-modulated signa! is applie 1 to the I.F. amplifier and the output of the amplifier is fed to the Y-plates of the oscilloscope to deflect the spot vertically. Horizontal deffection of the spot is obtained in the usual way by using the internal timehase of the oscilloscone. If the frequency modulation of the imput signal is synchrenised with the timebase of the oscilloscope, then a träce will

## 407 <br> BULATOR ".M. I.F. Alignment

VISUAL DISPLAY OF I.F. AMPLIFIER AND DISCRIMINATOR CHARACTERISTICS

By R. E. F. Street



Fig. 2.-The drilling dimensions of the chasshe.


Fig. 3.-The underchassis wiring and component layout.
be formed on the face of the tube which represents the amplifier characteristic.

I.F. response of an F.M. receiver at limiter grid.

## The Principle of Operation

In this instrument, a signal is generated at a frequency of $10.7 \mathrm{Mc} / \mathrm{s}$ - the optimum F.M. I.F. and this signal is frequency-modulated by a signal derived from the timebase of the oscilloscope. Early methods of obtaining frequency-modulation employed variation of the inductance of the oscillator producing the $10.7 \mathrm{Mc} / \mathrm{s}$ signal, or the use of a reactance valve circuit, which acted as a voltage-dependent capacitor across the tuned circuit of the oscillator. However, the circuit used in this unit employs a silicon-junction diode as a voltagedependent capacitor. When such a diode is biased in its reverse direction, a "depletion" layer is formed which acts as a dielectric of a capacitor formed by the two halves of the diode. The width of thïs layer, and thus the capacity of the junction, may be varied by altering the reverse bias across the diode. This is the principle of operation of the unit described.

It might be thought that the use of the $50 \mathrm{c} / \mathrm{s}$ mains voltage as a modulating source would make for a simple instrument, but using such a high sweep frequency would give a misleading set of curves-the tuned circuits would not have time to respond to the rapidly changing frequency of the input signal.
(Continued on page 430)

AIDIING CDMMUNICNTIONS HEATUIRES

## CHOICE OF VALVES AND STAGES

(Continued from page 342 of the August issue)

$\mathcal{G}$

By F. G. Rayer

OOD results can he nhtailled with the older type of value. such as those with octal bases. and many highly valued communications receivers have such valves. But when a receiver is being modified. or stages are to he added, it may be worth while using modern minialure valves. which will save space, and in some cases may give improved results.

## Volue of Modifications

If octal valves are to hand, or already fitted. it will not be worth while to replace these by miriature values. in some stages. This patticularly applies to andio amplifiers. Fiven in eatliet stages. an anticipated improvement from suhationting an old type valve by one with higher gain may scarcely be realised. This is hecause gain will almost always be reduced to some extent by the AVC circuit. so that the extra gain of the newer valve is not used.

It is in the early stage or stages in particular (e.g.. R.F. amplifiers, and mixer or fiequency changen that modern valves with a lower noice level can best be used. Even here. it musi be remembered that the noise level referred to is that
such as the 6K7 (or cquivalents CV1941, CV1943, FF39, OM6. Wh3 etc.) in R.F. and J.F. stages. Ihat such receivers are still in regilar use indicates that these valves can give good results. With then will often he found the $6 \mathrm{~K} 8,6 \mathrm{Q7}, 6116.6 \mathrm{Vb}$, and similar oetal valves.

Receivers of later design tend to use more modern valves. such as the $\mathrm{nSG} 7,6 \mathrm{BAt}$. UAF42, 6HJt. ctc.. in R.F. and I.F. stages. Frequencychangers. converters. mixers and oscillators are also of more modern lype, such as 6U8, UCH42, 6A18. 6All6. 6C4, etc.

In later stages. miniature valves are also generally used, bil some of these are equivalents of octal types. and are only fitted to save space and give uniformity. For cxample, a GRWG may be besed instead of a GV6. and except for being a miniatare is identical, for audio purposes.

To ohtain hest realls with the, moderm valve rypes, some resistor values may require changing. The circuits given here, and in other articles in this series, may he used with either octal or miniature valves. When adapting and improving a receiver, it should he emembered that a double valve can atten he infroduced. to perform more that one function. Fine example, a twin triode may act as A 1 : annlifier and BFO . Or such a value could he used for a $100 \mathrm{kc} / \mathrm{s}$ crystal marker. and audio or other purposes.


Fig. 13.-This circuit shows a 6BJ6 R.F. stage,
generated inside the set. and not that picked un by the aerial from external sources. which will often predominate.

The valve types favoured in commercially built communications receivers are a good guide to those which may he used when adapting or huilding a recciver. Many renutable communications receivers of older type. still in use, cmploy valves

Sk to 50 k . If the value is too sinall volume will he too great, even at minimum setting. But a very large valuc will give abrupt control of volume. A vatue of ahnut 20 k can thus often he fitted. The fixed resistor from cathode to H.T. positive ( 100 k in Jig. 1.3) helps to give more uniform control.
If a receiver of fairly simple type is in view. quite good results are possible on the lower


Fig. 14 (above).-A $6 B A 6$ valve in an l.F. stage.

Fig. 15 (right).-An E6F82 mixer-oscillator stage.
frequency amateur and S.W. bands, without the use of an R.F. stage. It is usual, howevor, to have at least one R.F. stage, as this reduces second channel interference, and improves sensitivity and signal-to-noise ratio.

## 1.F. Stages

When an ordinary type of superhet is being improved, it is usually fairly easy to add an I.F. stage. This will give considerable increase in selectivity and sensitivity. Such a stage is particularly easy to add in a receiver originally having only one I.F. stage. If the receiver already has two intermediate frequency stages, much more care is necessary to avoid instability. Complete screening, and decoupling of AVC and H.T. feed circuits, will be required.

A typical I.F. stage is shown in Fig. 14, using a high slope valve. Valves such as the 6 K 7 may be used, with a $220 \Omega$ bias resistor. and 47 k screen grid resistor. The additional I.F. transformer must, of course, be for the same frequency as those already fitted.
If the I.F. amplificr has
more than one valve it is useful to fit an I.F. gain control, which can be arranged in a similar manner to that in Fig. 13.
A crystal filter may be incorporated in an I.F. stage, and this will be dealt with later. A simple way of securing additional selectivity is to use two I.F. transformers in the coupling between I.F. stages. The secondary of the first transformer is connected to the primary of the second transformer through a very small capacity.

## Frequency-Changer or Mixer

Many popular sets have a triode-hexode frequency changer, such as an octal 6 K 8 , or miniature 12AH8. Older communcations receivers often have a valve such as the 6 K 8 operating as mixer only, with a separate oscillator valve. These methods can work well in general coverage receivers.

A mixer-oscillator stage using an ECF8: is shown in Fig. 15. A similar circuit may be used with a separate oscillator valve, and a 6 C 5.6 C 4 , 6 AMo . or similar valve may then be used as oscillator.

Fig: 15 also shows connections for a dipole aerial-the coil primary is not earthed, but twin sockets or terminals are provided, for the twin feeder. Such an aerial will usually give hetter results. especially in reducing untunable background noise from external sources.

Giood results can be achieved by using the ECF82 as the tirst slage. and this has the merit of simplicity. In a larger or more ambitious receiver, a 6BA6 R:F: stage, followed by the ECF82 stage, will allow of a very good performance.

C 1 and C 2 are sections of the usual gang condenser. The padder condenser $P$ should be of the correct value for the band in use. Several wavebands will normally be provided, as already explained.

## Combined Valves

When high sensitivity is required from a super-- het with few stages. a grid detector may he used, instead of the nore usual diode. Such a detector may be easily overloader, hut can give a vely good output with weak signals.


Fig. 17.-An AVC, detection and audio amplification circuit using a 6ATb.

A detector of this kind is shown in Fig. 16. Here, a double-triode has been used. and the second triode section is eniployed as BFO. The coupling condenser Cl is of very small capacity, such as can be made from twisting together insulated connecting wire for a short distance. Octal valves such as the 6SL7 or 6SN7 may he usefully employed. Full details of BFO oscillators have been given. The BFO circuit is tunable $1 \mathrm{he} / \mathrm{s}$ or so either side the intermediate frequency, C3 being used for adjustment. It renders CW morse audible, and is switched off for voice reception. Miniature

Fig. 18. - This output stage could use any of a number of volves: 6 V 6 .

B6W6, 6AQ5, etc,
valves, such as the 12A17, are equally suitable.

If a BFO is not wanted, or is already available. a $t w i n$ triode may be used as grid detector and A.F. amplifier. Alternatively, one triode section can furnish a $100 \mathrm{kc} / \mathrm{s}$ marker signal, as previously explained in an carlier article. Eiome receivers empley a twin tiode for BPO and audio amplitier.
Fig. 17 shows suitable values for a 6ATG, employed for detection. A.F. amplification. and automatic volume control. If a miniature valve is not wanted. a $6 \mathrm{Q}^{7}$ octal type may he used instead. An AVC in/out switch may he wired from the AVC, line in chassis, to permit manual control in R.F. and 1.I. stages, and prevent the automatic volume contral circuit trying in follow morse. A double-diode-triode stage of this kind is very pumblar indeed. The II.T. positive feed may be decoupled hy means of a 33 k resistor and. \& $\mu \mathrm{F}$ or similar condenser. This will slightly reduce hum, if II.T. smoothing is a little inadequate, and is generally worth while.
When phone reception is in view. the output of: the triode will offen be adequate. Alternatively, a second small triode can he used. Or a 6 H 6 ori other double-diode may he fitted. for detection and AVC and a double-triode to provide two stages, of A.F. amplification.

## Output and Rectifier

Very many communications receivers, in modern sets of high quality, employ a single heaing etrode or similar valve such as the 6 BW heam. haQs. These are miniatures, and simitar excert that the 6AOS is intended for up to 250 V encept and has a miniature 7 -pin hase. compared with. the 9 -pin of the fBWh. If octal valves are preSerred. a GV6 may be employed. with mo change in values or performance. Suitable values for any of these valves are shown in Fig. 18. Other valves may, of course, be used instead. If so, the hias resistor should be chosen to suit.
The existing power supply circuit of the receiver can probahly remain unchanged. especially if transformer and rectifier are of fairly generous rating. If the H.T. rating of the transformer is rather small. it may be necessary to replace the output valve by a more cconomical lype. for example, a 6BW6 might be replaced by a 6AM5, and this would save roughly 30 mA of H.T. current, which would be available for extra I.F. or other stages. The 6AM5 requires a 6802 bias resistor.

Club News


## REPORTS OF CURRENT ACTIVITIES

BRADFORD RADIO SOCTETY
Hon. Sec.: M. T. Powell, G3NNO 28 Gledhow Avenue, Roundhay, Leeds 8.
At the Annual General Meeting it was decided to change the name of the society from "Bradford Amateur Radio Society" to "Bradford Radio Society". On May 23 rd members visited the Granada TV studios in Manchester, and on June 6th they visited Amplifier Design and Corport. G3LZW gave a ralk on "Audio
Amplifier Design and Construction" on July 2 sth.
All meetings conmence at $7.30 \mathrm{p} . \mathrm{m}$. Slow Morse classes, if
previously arranged, are held before meetings.
Future Event:
Sepiember 12th-The first meeting of the new session.
DERBY AND DISTRICT AMATEUR RADIO SOCIETY
Hon. Sec.: F. C. Ward, G2CVV. 5 Uplands Arenue, Littleover, Derby.

The Society's Golden Jubilee Year celebrations are continuing, and on July 2nd the second Two Metre Field Day was held. A Surplus Sale was held on July Sth and on the 12th a Direction Finding Practice Run.

Future Events:
August Sth.-"Fifty Years of Radio"-an exhibition at the Derby Ar Gallery-will be held for three weeks until August 26 th. August $13 \mathrm{th}-A$ Mobile Rally at Rykneld School.
LICHFIELD AMATEUR RADIO SOCIETY
Hon. Sec.: T. L. Painier, G3NEO, Lyndhurst, 98 Gaia Lane,
The Society meets on the first Monday and third Tuesday of every month, at the King's Head, Lichtield.
LLANELLY AND DISTRICT AMATEUR RADIO CLUb
Hoo. Sec.: H. J. Hughes, 4 Pen-y-morfa, New Dock, Llanelly.
Lectures for the RAE were continued each Thursday up to the end of May. A highly successful "Ladies' Night" was held
on May 18 th.
During Technical Training Week, commencing May 30th, a club transmitter, under the call sign GW3LL U/A, was set up at the Drill Hall, Llanelly.

## EUTON AND DISTRICT RADIO SOCIETY

Bon. Sec.: D. Bavister, 70 Crawley Green Road, Luton, Bedfordthire.
The Society is organising a Mobile Rally for August 20th at Stockwood Park, Luton.
Meetings are still held at Surrey Street School every Monday rilght at 8 p.m.

## MTTCHAM AND DISTRICT RADIO SOCIETY

Hon. Sec.: M. Pharaoh, G3LCH, 1 Madeira Road, Miteham.
Recent meetings included one where various members brought along a piece of equipment and gave a shor-descriptive talk thbut it. On Friday, June 30th, there was a discussion on Club dffairs and outside activities.
On July 22 nd members operated an exhibition station at the Mitcham Horticultural Show. The Society operated a similar station at the Show last year.
NORTHERN HEIGHTS AMATEUR RADIO SOCIETY Hon. Sec. : A. Robinson, G3MDW, Candy Cabin, Ogden, Halifaz,

On July 26th the Society held-an informal meeting.
Future Event:
August 9th-A discussion on the Scout-Jamboree-On-The-Air, NORTHERN MOBILE RADIO RALLY
Hon. Sec.: J. Charlesworth, G3IJC, 23 Craven Lane, Gomersal, Loeds.

The Rally was held in fine weather and attracted approxithately 900 visitors of which 80 were mobiles, which with 160 non-mobiles, made a total of 240 cars. The majority operated on 160 m , although a few 2 m and All-band operators were present.

## PADDINGTON AND DISTRICT AMATEUR RADIO <br> SOCIETY

Hon. Sec.: N. A. Lambert, G3LVK, Beauchamp Lodge Settlement, 2 War wick Cresceat, London, W.2.
The Society continues to meet on Wednesday evenings at 7.30 . Recent activities have included a sale of surplus equipment and talks by Mr. Alban (G3JEA), and Mr. Legge (G3KNL). Mr. Kippin, G8PL, brought along a large number of pre-war QSL cards and certificates to show the members.
On Jane 24th the Society was active on G3PAD and also had an exhibition of home-made equipment.

## SHEFFIELD AMATEUR RADIO CLUB

## Hon. Sec.: D. R. A. Hill, 16 Tylney Road, Sheffield 2.

The two technical lectures will be held during October and December at the Dog and Partridge Hotel, Trippett Lane Sheffield 1. They are to be held on the second Wednesday of
The October lecture will be on High Power Transmitters and will be given by Dr. Kaiser Department of Physics, Sheffield University; and the December lecture will be given by Mr. Lyon, and is entitled "New Receiver".

## SOUTH YORKSHIRE AMATEUR RADIO SOCIETY

Hon. Sec.: E. Brailsford, G3PAF, 15 Ayrıome Walk, Cantley 4,
Doncaster.
The Society continues to increase it membership, and is now on the air as G30WK. Members recently saw two Mullard films at the Club room.

## An Intercom Amplifier

(Continued from page 405)
Pin 1 on each valve base should be carthed to chassis.

## Testing

When construction is complete and the wiring has been checked against the diagram, test with a meter between $C 7$ and chassis that there are no shorts in the H.T. circuits. If all is well, apply power and check that the proper voltages appear at the valve electrodes. R 10 can then be. connected to the output transformer and if this causes oscillation, the connections to either the primary or secondary of the output transformer should be reversed to make the feedback negative. Orientate Tl to the position of minimum hum and bolt it to the chassis; hum should then be inaudible Gin. from the speaker when the primary of the input transformer is short circuited.

## Operation

The external lines should be unshielded and if an earth connection is used, and it is not essential, it should be to the amplifier chassis only. In this way, interference which may be present is picked up on both transmission lines equally and is cancelled out in the primary of TI. Ordinary lighting flex is very satisfactory for distances up to about 50 ft but for longer runs, a cable of lower D.C. resistance is desirable. The signal source may be a moving coil microphone but a small moving coil speaker is equally satisfactory and much cheaper.

## Talk-back Facilities

For intercom service, talk-back facilities will be required and may be arranged quite simply with the circuit of Fig. 3. The switch should be a double-pole change-over component preferably of the key type, but, if this is not obtainable, an ordinary rotary switch can be modified by removal of the locating device and the addition of a spring-loaded arm as shown in Fig. 5, so that it will automatically return to the "Listen" position after use. Communication to any number of distant points can of course be established by including a multi-way switch in the transmission line at point $X$ in Fig. 3.

MEDIUM, LONG WAVE AND TRAWLER BAND EXTENDING TO* 60 METRES WITHOUT COIL CHANGING $350 \mathrm{Mm} \times \mathrm{ClOl}$ 's push-puli output Transistors. Powerful magnet $3 i n$. high grade speaker. Miniature pushpull eransformers. This is top pertorming recoiver. Nearly 30 stations listed in one evening including Luxembourg loud and clear. A pleasure to listen to. FERRITE ROD AERIAL. All parts sold separately, including pale blue gleaming polystyrene case with duo-diffusion grilles in red. Uses 9 volt battery. Soskets for car aerial.
Total building cost $\mathbf{1 6 . 1 9 . 6}$ P.P. 2/6. Size $6 \frac{1}{2} \times 4 \frac{1}{2} \times 1$ tin. "Agreeably surprised with Trawler Band recention. Luxembourg as loud as local. Your easy build diagram helped a lot . . . my hrst attembt."-H.S. Penzance, Comwall (poor reception area).
"Super car radio."-l.B.B.V., Liverboal. ALL PARTS SOLD SEPARATELY

## TRANSONA-4

(4'Ediswan Transistors, plusiz Drodes)
Ministure speaker. FERRITEROD AERIAL. MW:LW and Trawier Band Soverage down to Bo metres On test tuned in many stations. This sensational new 'radio is simple to build with our easy. build plans. Handsome pocket
 ease.

- May be built ior 65/- P.P.-2/6


## NEW! <br> POCKET RADIO

(5 Ediswan Transistors and 2 Oiodes)


MEDIUM, LONG. WAVE AND TRAWLER BAND TO 80 METRES. Designed round super sensitive territe rod aerial and 3 in. speaker. Home and Continental stations at your finger tips. Attractive 2-tone Docket size case in gleaming polystyrene. No astial required. Many stations listed test receiver. Easy-build plans for beginnars. Powared by $4 \frac{1}{2}$ volt battery.

Total cost of parts required. 24.|9.6 P.P. $2 / 6$ ALL PARTS SOLD SEPARATELY

NEW PUSH-PULL FIVE
(MW/LW and TRAWLER BAND)
(5 Ediswan Transistors)


Now greatly improved. Sensitive FERRITE ROD AERIAL. Pale blue polystyrene case with speaker grilles in red. Volume/sensitivity control. Miniature push-pull transformers. 350 Mw XClol's in pushpull. Sensitivity R.F. stage for station selection.
local building cost 5.7 .6 'P.P. $2 / 6$ ALL PARTS SOLD SEPARATELY

## Push-Pull Pocket Six

MEDIUM AND LONG WAVES AND 400 TO 750 METRES WITHOUT COIL CHANGING. Sensitivity of a superhet, tonal quality of a TRF. Volume control. Tuning condenser. Latest type switches. Handsome two tonepocket case. Ferrite rod serial. 3 in , quality speaker. Easy build diagrams. 6 Transistors (ineluding Ediswan and Semiconductors) plus 2 diodes.

> All parts cost \&5. 19.6 P.P. $2 / 6$ ALL PARTS SOLD SEPARATELY

## New!

Easy Build Three Radio
(3 Ediswan Transistore plus 2 Diodes)
Easy to build, easy to operate. This transistor radio operates over the M.W. and L.W. extending down to 80 metres without coil changing. Ferrite rod aerial, volume-sensitivity control. Sonotone high fidelity miniature arpiece or minia
ture speaker.
May be built for
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# SIGNAL GENERATOR OUTPUTS 

(Comfinued from page 315 of the Alugust issue)

By R. Brown

7
can he nutput imnedance of a cignal generator external impedances.

Supposing the required nutput impedance is greater than the actual signal generator
imnordance. the signal generator output impedance can he increased to the new value by connecting a resistance $R B$ in series with the nutnut lead. $R$ g must have a value such that $R$ R $+R_{A}$ is equal to the required output impedance. Under these

| Output Connections | Rfotiren Oitpyt Impedance Ro or Zo | Value of Extra Resistor Rв | Npw Source EMF | Voltage Developrd. Across Load R |
| :---: | :---: | :---: | :---: | :---: |
|  | Resistive Ro>RA | Ro-RA | E | $\frac{R}{R A+R_{B}+R} \cdot E$ |
|  | Resistive <br> Ro<RA | $\frac{\text { Ro. } \mathrm{RA}^{\text {a }}}{\text { RA-Ro }}$ | $\frac{R_{B}}{R_{A}+R_{B}} \cdot E$ | $\begin{gathered} \frac{R \cdot R_{B}}{R_{1} \cdot R_{B}+R_{A} \cdot R_{B}+R \cdot R_{A A}} \cdot E \\ \vdots \\ \hline \end{gathered}$ |
|  | Resistive and reactive (Ro+iX) Ro $>$ RA | Ro-Ra | E | $\frac{R}{\left.\sqrt{\left\{(R A+R B+R)^{\prime}+X^{\prime}\right\}}\right\}}$ |
|  | Resistive and reactive (Ro+iX) Ro<RA | $\frac{\text { Ro.RA }}{R_{A}-R_{0}}$ | $\frac{\mathrm{Ra}}{\mathrm{RA}+\mathrm{RB}} . \mathrm{E}$ | $\left.\sqrt{\sqrt{y}\left\{\left(R^{R}+\frac{R A \cdot R B}{R_{A}+R_{B}}\right)\right.}+X^{B}\right\}$ |
|  | Resistive $\mathrm{Ro}=\mathrm{RA}_{\mathrm{A}}$ | - | E | $\frac{R}{R_{A}+R} \cdot E$ |

The connecting cable $X-X$ should have a characteristic impedance equal to RA, $Y$ - $Y$ should have a : characteristic impedance equal to RO. (or ZO).

Fig. 4.-Altering the signal generator output impedance. For convenience $X$ may be connected at load end of $Y$. $Y$, which should then hove a characteristic impedance equal to RO.
conditions the source EMF will have the same value, E , as before but the voltage developed across a load, $R$, will now equal:

## ER

$$
\begin{equation*}
\mathbf{R}_{\mathbf{A}}+\mathbf{R}_{\mathbf{B}}+\mathbf{R} \tag{1}
\end{equation*}
$$

A reduction in the signal generator output impedance cai be achieved by connecting a suitable value resistor, $\mathrm{RB}_{\mathrm{B}}$, directly across the signal generator output. The value of $\mathrm{R}_{\mathrm{B}}$ is chosen so that the parallel combination of R , and RA ,
connecting the signal generator to the receiver via a network which has an impedance equivalent to the aerial impedance. A standard network, called a dummy aerial, is shown in Fig. 5. This, with the output impedance of the signal generator, effectively simulates a broadcast band aerial.

## The Connecting Cable

The signal generator is usually connected to the receiver with a length of coaxial cable. This cable does not need to be terminated in a resistance equal to its characteristic impedance. It must, however, have a characteristic impedance equal to the signal generator output impedance.

That this is so can perhaps best be seen with the aid of Fig. 6. This shows a signal generator, to the output of which is connected a length ' 1 ', of coaxial cable. The far end of this cable is open circuited. It has a characteristic impedance equal to Ra. Now, it can be shown, that if such a cable is "loss-less", then the voltage Vi developed across the input end will be given by:

$$
\begin{equation*}
V i=E \cos 81 \tag{5}
\end{equation*}
$$

where B is the phase constant ( $2 \pi 1 \lambda$ )' of the cable.
It can also be shown that the voltage Vo at the end of such an open-circuited line is given by:

$$
\begin{align*}
\mathrm{Vo} & =\mathrm{Vi} / \cos \mathrm{B} 1 \\
& =\mathrm{Ecossl} / \mathrm{cossl} \\
\text { Therefore, } \mathrm{Vo} & =\mathrm{E} \tag{6}
\end{align*}
$$

In other words the source EMF is still E volts even though the cable is unterminated.
If we now look back along the cable from the far end we see a line of characteristic impedance Zo terminated in a resistance RA. But RA=Zo. So we are looking at a cable which is correctly terminated. The impedance we seé looking back from the far end of the cable is therefore RA.
Thus provided the cable does have the correct characteristic impedance, the receiver, or other


Fig. 7.-The cable output voltage for on open circuited cable which has a characteristic impedance equal to $2 R A$.
equipment. to which the signal generator is connected will still see a generator, of $E$ volts in series with a resistance RA.

- (Continued on page 422)


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By J. B. Dance

# FROM 

$\mathcal{J}$APE recorders do not normally have built-in radio receivers. When one wishes to record a radio programme, it is, therefore, necessary to feed a signal from a receiver into the input of the tape recorder through a length of cable which may have to be fairly long. It is preferable to use coaxial cable.


Fig. 1.-The cathode follower detector.


## Signal Supply Point

It is common practice to take the signal from the external speaker terminals of the radio receiver or (if no such terminals are fitted) the signal may be taken from the output side of the speaker transformer. Although a low impedance output is conveniently obtained by this method, it necessarily means that any distortion in the receiver output valve or in the speaker transformer will be present in the recording. Most radio receiver distortion usually occurs in the output stage unless the receiver is an expensive one.

On the other hand, it is not wise to connect a length of coaxial cable to the detector output or to the anode of an audio amplifier which is in the radio receiver (via a D.C. blocking condenser), because the capacity of the cable will severely affect the high frequency response.

## Cathode Follower Detector

If a radio receiver is being constructed so that it can be used to feed a tape recorder, a cathode follower detector may conveniently be used. This circuit, which is shown in Fig. 1, has a low output impedance and can be used to feed a tape recorder through a long length of coaxial cable without high audio frequency attentuation. The circuit


Fig. 2. -The circuit on the right is a cathode follower audio stage which may be added to a receiver the existlng circuit of which is shown in the left-hand diagram. Point " $X$ " should be connected to one of the points " $A$ ", " $B$ " or " $C$ ".
has the additional advantages that it does not appreciably load the previous tuned circuit and maximum selectivity and gain are therefore obiained. The distortion given by the cathode follower detector is very low. Almost any smalt triode or triode connected pentode, e.g. the 6C4 triode, can be used -as the detector valve in the Fig. I'circuit.

## Cathode Follower Stage

If it 'is.' required to feed a tape recorder from an existing receiver, the substitution of a cathode follower detector for the common diode detector would involve re-alignment of the receiver. This can be avoided by the use of an additional cathode follower audio stage in the receiver.
Assuming a diode detector is used in the usual double diode triode circuit shown in Fig. 2, the audio output may be taken from one of the points marked $\mathbf{A}, \mathbf{B}$ and $\mathbf{C}$ in the circuit and fed into the point marked $X$. The component values shown in the left-hand circuit of Fig. 2 are typical, but will vary slightly from receiver to receiver.

If the point $X$ is connected to the point $A$, considerably more amplification will be obtained than if it is connected to the point B or point C . The receiver volume control will affect the tape recorder input if the connection is made to point $A$ or $B$, but it is probábly more convenient to connect point $X$ to point $C$, so that the receiver volume control does not affect the recording level.

## Double Trlode Circuit

Those readers wishing to use the double triode ECC82 (or the 6SN7 or 12AU7) may find the Fig. 3 circuit convenient. The first triode is an ordinary amplifying stage, which is directly coupled to the grid of the second stage. The positive voltage applied to this grid (about 100) is counteracted by': a slightly greater positive voltage of the cathode owing to the flow of the anode current throught the large cathode resistor. The difference

[^5]

Fig. 3.-A double triode as amplifier and "cathode follower.
between these voltages provides the bias for the second triode. The input can be taken from points B or C of Fig. 2. The audio voltage at point A is too large for connection to the input of the Fig. 3 circuit.

With the circuits shown, the signal can be fed into a very long lenght of coaxial cable, and hence to the recorder, without the high frequency response being noticeably affected. The circuits must be placed in the radio receiver and not in the tape recorder. When the circuit of Fig. 2 or Fig. 3 is used, the lead from the audio take-off print ( $\mathrm{A}, \mathrm{B}$ or C ) to the additional valve circuit stiould not be more than about three inches long -shorter if pussible. Care should be taken to ensure that there is only one earth connection between the receiver and recorder; this is the outer comection of the coaxial cable.

## SIGNAL GENERATOR OUTPUTS

## (Contimued from page 415)

There will, of course, be standing waves on the cable (except when the luad equals RA), but provided the cable introduces negligible losses. this will not cause any error. The cable will not normally introduce losses, for the signal generator will be close to the equipment under test, and the cable will be very short.

## Incorrect Cable Impedance

The incorrect reading of the signal generator attenuator, and an incorrect connection to the equipment under test can certainly result in errors in measurements. The cause of these errors is often difficull to tind. A graph is shown in tig. 7 giving the variations in the cable output voltage with length, when the cable has a characteristic imperdance equal to $2 R$ a. The output impedance seen by the equipment under test will also be wrong with this condition, and it does not fueed math imagination to picture the completely incorrect and probably battling results which could be achieved. Provided reasonable care is taken, however, one can usitally be sure what output voltage the signal generator is Eiving, and what its output impedange is.

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| 5 Y 30 | b/9 | 6U4GT | 016 | 6158 | 11/- | ECF82 | 8/8 |  |  |  |  |  |
| 573071 | $8 /-$ | ${ }^{60}$ | 5/6 | buay | 9/- | ECi21 |  |  |  |  |  |  |
| 5 C 46 | 8/8 | 6 VbG | 6/6 | 18587 | 18). | ECH35 | 9/6 | PCesy |  | U324 |  | \% |
| 0.4 | 8/8 | 6x 6 | $6{ }_{6}^{6}$ | 307(A) | 6/6 | ECH4L | 816 | PCFO |  | -is89 |  |  |
| 6A8G | 8/8 |  | $5 / 8$ | 955 | $8 / 8$ | ECLEA |  |  | $7 / 3$ | U4us |  | 18 |
| C7 | 4/8 | OY0 | $7 / 8$ | 456 | 2/6 | ECL8\% |  |  |  |  |  |  |
| 65 | $9 / 8$ | 7B6 |  | -0 | 4/- | Eclss | $12 /$ |  |  | ${ }^{\text {a }}$ AF4d |  |  |
| ${ }_{\text {K }}{ }^{\text {c }}$ |  | 787 | $7 / 9$ |  |  | EF24 | $7 /$ | PCL. 84 | 7/6 | Cbed |  | 1 |
| dals | 810 | \% | $7 / 8$ | aT |  |  | $8 / 8$ | PEN | 4/8 | 4 BC |  | 1 |
| 6AME | 1/. | [ Hz | $7 / 10$ | 888 |  | Fipso |  |  | 78 | UBE |  | 8 |
| bavs | 8/- |  |  | B6is |  | EFEḃ | 1 |  |  | UCB |  |  |
| 6aTb | 878 |  | $7 \%$ | cblal | $21 /-$ | Er'zy | 8/8 | PLis | 10/6 | UCH28 |  |  |
| \%AESO |  | 1ucl |  | cchss | 14/9 | ER91 | $8 /-$ | PL; 8 | 16/6 | Wer |  | $8 / 8$ |
| UBAH | $8 \cdot$ | 10 F | $5 / 9$ | CY 4 |  | EL | 4/816 | 181 | $8 / 8$ | UCLEL |  |  |
| HBEC | 5/8 | 10LDill | 14/6 | D63 | 1/6 |  | 4/8 |  |  | Cl |  |  |
| GBG6E 6BW\% | 12/8 | 10 P | 9/- | DA | 2/8 | EL38 |  |  |  |  |  | 8/8 |
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| -0 | 4/8 | 12act | $7 / 6$ | DF | 9/9 | ELas |  |  |  | UF66 |  |  |
|  | 8/8 | 12ar? | 5/6 | Dr91 | 8/2 | EL43 |  |  |  | UF59 |  | \% |
| 4 CD 8 C | $21 /-$ | L2AU7 | 1 | DF92 | $3 / 8$ | ELst |  |  |  |  |  |  |
| ${ }_{80 \mathrm{CH}}^{8}$ | $8 / 8$ | 12AY7 | 8/9 | - Fg 98 |  | EL91 | 4/6 | Y234 | 9/8: | LL48 |  |  |
| 806 8151 |  |  |  | DH7\% |  | EM34 | $8 / 6$ | 818 | 11/- | CL34 |  | 716 |
| 6FEG | $8 / 8$ | 12 K 8 GT |  | DEst |  | EM80 |  | K19 | 11/4 | Ulva |  | - |
| 8 EF | $3 /-$ |  | 11/- | D691 |  | EM |  |  | 16 |  |  |  |
| ${ }_{6} \mathrm{OF}^{6} 14$ | 8/9 | 12470T | $5 /$ | DE9\% | $7 / 8$ | EM8̇ |  |  |  |  |  |  |
| ${ }_{6}^{6 F 14}$ | 9/8. | ${ }^{1285}$ |  | DE96 |  | EN:1 |  |  | 8 j | UY41 |  |  |
| ${ }^{6815} 8$ | 9/8/-1 | 12SK7GT |  |  |  | EY61 |  | ${ }^{\text {U22 }}$ | $8 / 8$ | UY85 |  | $8 / 8$ |
| 8550 | 2/18 | 128N7GT |  | DLGI |  | Stmbill |  |  |  | V H 10 F |  |  |
| ${ }^{0.550}$ | 8/8 |  | 8/6 | Lisy |  |  |  |  |  |  |  |  |
| ${ }_{6 J 76}$ | 5/8 | 1303 | 71 | Dis3 |  | EZ41 |  |  |  |  |  |  |
| 6.J7t 6 KOGT | 7/8 | ${ }^{1487}$ | $28 / 6$ | DLyt |  | Fizso | dj- | U33 | 14/- | x 63 |  | $1 / 8$ |
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# Converting the TRF Transistor Four to L.W. Tuning 

A SIMPLE MODIFICATION TO THIS P.W. SET TO ALLOW RECEPTION ON THE LONG WAVE BAND.

$\mathcal{J}$HE four transistor TRF receiver featured in the June 1960 issue of Practical Wireless was described as simple to construct and therefore particularly suitable to the newcomer to transisitor circuits. For simplicity, medium waveband tuning only was included, and this arrangement is entirely satisfactory for readers living equidistant from the transmitters covering the Light and the Home programmes on the medium waveband. However, in areas where it is the usual practice to receive the Light programme on the long waveband, medium waveband tuning only restricts the usefulness of the receiver, and details are given here of how best to modify the original design, in order to accommodate both wavebands. The modification necessitates introducing a four-pole, two-way, switch into the circuit, which may be the circular type ( 1 i in . in diameter) and also an additional L.W. R.F. transformer.


Fig. 1.-The arrangement of the second tuned circuit.
The L.W. Coil used is a Teletron HLX.

By D. SAULL

In the first method, where both coils are wound on the same ferrite rod, the medium-wave coil forms parts of the winding of the long-wave coil. When the wave-change switch is turned to the medium-wave position, the long-wave section of the coil is left floating. This section of the coil has an inductance of about 2.2 mH and requires only a 10 pF parallel capacitance to resonate at $1 \cdot 2 \mathrm{Mc} / \mathrm{s}$, which is the wavelength of the Light programme on the medium waveband. Should $:$ this section of the coil, when not in circuit, so resonate -being tightly coupled through the ferrite rodi to the medium-wave tuning section of the coil-it will result in heavily loading the medium-wave coil and will render the $1.2 \mathrm{Mc} / \mathrm{s}$ tuning section inoperative. The self capacity of the wave-change switch will be about 3 pF , and the single-layer wound aerial coil will also possess a little self capacity-hence the stray capacity of the wiring in the circuit must be kept to an absolute minimum if trouble $\dot{i s}$, to be avoided. This may be achieved by keeping the connecting lead from the top end of the long-wave coil to the switch as short as possible. To achipve this. the wave-change switch should be mounted directly beneath the ferrite rod-the switch control spindle may then extend to the rear of the receiver cabinet.
The Method of Winding the MW. and L.W. Coils on the tame Ferrite Rod
The dual-wave coil consists of 150 turns of 34 s.w.g. double silk covered wire, close wound in a single layer on the forther that should be a sliding fit on the two ferrite rods; the coil is tapped as shown in Fig, 2. The former is made by winding a strip of thick paper, two inches wide, three times round the ferrite rods, adhesive haking first been applied to the surfaces in contact. A piece' of similar paper wrapped round the ferrite rods beforehand

## The Construction of the Aerial Colls

There are three methods of constructing the aerial coils: winding the coils for both wavebands on the same ferrite rod; winding the medium-wave coil on the ferrite rod and winding a separate frame aerial to cover the long-wave tuning; or winding the long-wave coil on the ferrite rod and winding a separate frame aerial to cover mediumwave tuning. If either of the last two methods is used, the choice between them is determined by the weakest signal strength in a particular area-a frame aerial has greater sensitivity to weaker signal strengths.


Fig. 2.-The winding details of the dual-wave coil.
will ensure a little clearance, so that the finished former is free to slide on the ferrite rod when the under piece is removed.
The gauge of wire is not critical, and it is not important that double silk covered wire is used enamelled covered wire would function electrically. However, the best conditions prevail if the given specifications are adhered to- and the guide to the physical measurements, in Fig. 2, may then be followed.
rods adjusted to give the best results. This position will most likely be in the centre.

## Winding the M.W. and L.W. Colls Separately-One or a Ferrite Rod-the Other as a Frame Aerial

If the L.W. coil is to be wound as a frame aerial, the existing M.W. winding on the ferrite rods need not be disturbed. If the L.W. coil is to be on the ferrite rod, then wind the coil as in the first method, omitting the 10 -turn tap.
The frame aerial may be wound on the inside ot the back cover of the receiver cabinet; the size should be about $7 \frac{\mathrm{tin}}{\mathrm{in}} \mathrm{in}$, $10 \frac{3}{3} \mathrm{in}$ The coil is wound round fou: blocks, $\frac{3}{3}$ in. square, cut: from hardboard and cernented to the inside of the back cover of the receiver; four larger blocks. $\frac{7}{8}$ in square, also cut fróm, bard. board, are cemented on top of the former blocks, to prevent the wire from sliding. of (Fig. 4).

Fig. 3. (left)-The method of winding the coil on its former.


Fig. 4.-This shows the way in which the frame oerial is. wound on to the inside of the back of the receiver.
The taps on the coil are formed by making a loop in the wire and twisting together the pair thus formed: then carry on winding.
A convenient method to wind the coil is to cut a narrow piece of Sellotape to a size about tin. $x \frac{1}{2}$ in. to form an anchoring tab. Then wrap the extreme end of the tape round the wire and attach the wire to the former (Fig. 3a) by means of the remaining length of adhesive tape. Successive turns of the winding will hold the tape firmly in position. Likewise, slip a piece of tape, the same size as betore, underneath the wire (adhesive side uppermost) about 15 turns from the completion of the winding (Fig. 3b). When the final thita has been wound on, pull the extruding length of tape upwards and back over the top of the winding. A coating of varnish over the entire winding will hold it secure, but first remove the former from the ferrite rods, otherwise it may stick to the rods. The ferrite rod aerial may then be connected, and the position of the coil on, the


Fig. 5.-This shows the method of connecting the frame aerial in circuit.
For medium-wave tuning, the frame aerial.con sists of 12 turns of 34 s.w.g., d.s.c., wire tapped ai 4 turns: for long-wave tuning, 51 turns of the same type wire is tapped at 17 turns, connected as in
Fig. 5 .

## The Second Tuned Circuit

The arrangement of the second tuned circuit remains identical for the three methods; and is shown in Fig. 1.

The L.W. coil used is a Teletron; HLX.
Ascertain that connecting wires in the first and second tuned circuits of the receiver are not in close proximity, which would result in instability.

The method of aligning the receiver is the same as that described in the June 1960 issue, but in the lirst instance make sure that the 40 pF reaction capacitor is fully out and then slowly turn in. to increase the positive feed-back until the best results are obtained once the required signal is
heard.

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The Editor does not necessarily agree with the opinions expressed by his correspondents

Whilst we are always pleased to assist readers with their technical difficulties, we regret that we are unable to supply diagrams or provide instructions for modifying commerical or surplus equipment. We cannot supply alternative details for receivers described in these pages. WE CANNOT UR receivers described in these pages OVER THETELEUNOERTAKE TO ANSWER AUERIES PHONE. If a postal reply is required a stamped and addressed envielope must be enclosed with the coupon from page 111 of the cover.

## TRANSMITTER LICENCES

$\mathrm{S}^{1}$IR,-Surely we have not started that stupid business about low-power transmitter licences again? Mr. Dick's letter in the July issue seems to point to this. To own a "ham" transmitter licence is a privilege, not a right, and it seems only fair that one should earn this privilege by taking the G.P.O. examination. As for saying that only a simple test, if any, would be required, is ridiculous; the present test does not seem all that difficult.

Regarding his wish to abolish the morse test for inputs of under one watt, I should have thought that morse was the only way to achieve a reasonable range with this type of equipment. Moreover, if this test were abolished, there would be dozens of irresponsible people using ex-government transceivers, which, among other things, are by no means TVI-proof.-M. J. RfDMAN (Brighton).

SIR,-I have read with inferest the great deal of correspondence which you have published recently on the subject of novice licences.
It has been my ambition for the last three years to obtain an amateur transmitting licence, and at last I have taken the Radio Amateurs Examination; and am now striving to attain the morse sending speed necessary for the G.P.O. test. If, and when, I finally obtain my licence, I shall feel justified in using the frequency bands allocated to amateurs. I have been asked by many people studying science subjects what justification amateurs have in using parts of the precious frequency spectrum for their own pleasure. I then proceed to tell them that at least the amateur has a knowledge of morse, which might be valuable in an emergency.

If novice licences or special concessions to "phone-only" operators were permitted in this country, the people who have to defend the amateur's cause would have no argument for the continued existence of amateur bands.

It takes three years as an SWL to becone really familiar with amateur procedure, and anyone who cannot accumulate enough knowledge in that time to pass the RAE cannot be really interested in amateur radio.-D. A, Park (Bagshot).

## ITV SOUND

SIR,-The reception of TV sound on a radio receiver, as described by B. Quest (July issue) can sometimes be caused by the sound channel of a local TV station interacting with part of the lower vision side-band to form an intermediate frequency; similar to the. I.F. generated in a frequency changer.

Reception of such a signal can only be on a sensitive TRF receiver; not with a superhet. This effect disappears below $1 \mathrm{Mc} / \mathrm{s}-750 \mathrm{kc} / \mathrm{s}$, this being the separation between the furthest sound and vision signals, and it should disappear altogether when the vision carrier is not modulated. - $\mathbf{C}$. barnes (Cheshire).

## vintage models

S
IR,-I was not surprised to read in the July issue of P.W. a letter from Mr. A. V. Newman expressing his pleasure on hearing a friend's 1934 radio.
The pleasant tone of these vintage sets is a phenomenon not met in these days of "high fidelity".
1 would not care to open any discussion on the merits or demerits of high fidelity in these columns-it would rage on indefinitely-but. the pleasure derived from listening to music on the old faithfuls is almost entirely due to the absence of the more irritating top frequencies and the presence of a measure of "woofiness." from-the larger cabinets with their uncontrolled resonances. -W. J. Huntingard (Guildford).

## CORRESPONDENTS WANTED

SIR,-I am 15 years old and I am very interested in radio and TV construction and theory. I would therefore like to correspond with any other readers of the same age.-H. Johnstone (2 Fox Hill, Distington, Cumberland).

## DISHEARTENED SWL

SIR,-Being one of the "younger SWL's set", and after reading in several books that local hams are always ready to help SWLs, I decided to visit a nearby operator.

The books could not have been more wrong! He considered that SWLs were maniacs, forgetting that he was once one himself.
1 can only hope that not all amateur radio enthusiasts hold similar opinions.-E. L. W. (Cheshire).

## THREE FREE IBLUEPRINTS

WITH the October, November and December issues we are presenting free blueprints intended to teach the beginner the fundamentals of practical radio construction and also to enable him to build useful receivers for his own use. The final design will be a transistorised superhet tuner unit with an amplifier which may also be used with a battery-operated record player.

## EACH DESIGN WTLL INCORPORATE <br> ---inexpensive construction <br> -readily available components <br> -long and medium wave reception

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## A Wobbulator for F.M. I.F. Alignment

(Cominued from page 410)
The circuit of the instrument is shown in Fig. 1. A signal derived from the $X$-delecting plates of the oscilloscope is fed to VIA which is arranged in a cathode-follower circuit. The silicon-junction diode is connected from the cathode load of this valve to the tuned circuit of the oscillator, which is set to give a centre (unmodulated) frequency of $10 \cdot 7 \mathrm{Mc} / \mathrm{s}$. The voltage across the cathode resistor (R4) also provides the reverse bias necessary for the diode. The decoupling condensers C2 and C3 are siven large values to ertsure that the tirmebase


Fig. 4.-Details of the coil windings (on $\frac{1}{4}$ in. former with dust core).
waveform suffers negligible distortion in passing through VIA.

Although the second half of V1 is not used in the circuit, it is given a feed from the H.T. So that a current passes through it all the time that the unit is switched on: this avoids "cathode poison= ing " or loss of emission in VIB.

The $10.7 \mathrm{Mc} / \mathrm{s}$ oscillator is provided by V2A and the circuit values are so arranged that the freyuency of operation is $10.7 \mathrm{Me} / \mathrm{s}$ when there is no input to VIA. This centre frequency is set by altering the setting of a dust core in the former of coil lif.
(To be cominued)

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Resistors: ${ }^{\text {al }}$ |  |  |  |  |  |  |  |  |
| R2 IM | R 5 | 100k |  | 4.7 | R11 | $2170 \Omega$ |  |  |  |
| R3 $220 \Omega$ | R6 | 100k |  | 1 M |  | Ik |  |  | or 25k |
| Capacitors: |  |  |  |  |  |  |  |  |  |
| 61 0.25 F F 350VW Paper |  |  |  |  | C7 50pF Ceramic |  |  |  |  |
| C2 $32 \mu \mathrm{~F}$ | $0.25 \mu \mathrm{~F} 350 \mathrm{VW}$ Paper$32 \mu \mathrm{~F} 350 \mathrm{~W}$-Daly W3 13.10 |  |  |  |  |  |  |  |  |
| C3 $32 \mu \mathrm{~F}$ | $32 \mu \mathrm{~F} 350 \mathrm{VW}$-Daly W3 13.10 |  |  |  |  |  |  |  |  |
| C4 ${ }^{\text {C5 }} 1000 \mathrm{pF}$ | 1000pF Ceramic |  |  |  | CiO 0.01 hF Paper 250 |  |  |  |  |
| C5 1000 pF | 1000pF Ceramic |  |  |  | CII, CI2 16, $16 \mu \mathrm{~F} 350 \mathrm{VW}$ - |  |  |  |  |
| Valves: VIA | VIA and VIB |  |  |  | Daly W2/39/10 |  |  |  |  |
| Diode: DI SVCI GEC Semiconductor Division, School Sireet, Hazel Grove, Stock port, Cheshire <br> Chassis: 6 in $\times 4$ in. $\times 2 \frac{1}{2}$ in., approximately |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| Coll: This is wound on a $\frac{1}{4}$ in. internally threaded coil former and details are given in Fig. 4. A dust core is required. |  |  |  |  |  |  |  |  |  |
| Sundries: Two B9A valveholders, tag sirips, mains switch, wire, etc. |  |  |  |  |  |  |  |  |  |
| Transformer: Mains primary, 180 V to 250 V H.T. secondary, 6.3 V IA winding for heaters. (A "converter" transformer is very sultable.) |  |  |  |  |  |  |  |  |  |
| Rectifier: Miniature contact-cooled type-200-250V, 40 mA |  |  |  |  |  |  |  |  |  |
| Note that the value of RII may require to be altered to ensure that the H.T. line voltage does not exceed 190. |  |  |  |  |  |  |  |  |  |

$465 \mathrm{Kc} / \mathrm{z}$. SIGNAL GENERATOR. Total coul 15/- Uee B.F.O. Unlt 2 A 30038 ready made POCEET SIZE $24 \times 41 \times 1 \mathrm{n}$.
Bilsbt modifcations required, full tartructions supplied. Battery $7 / 8$ extra $69 \mathrm{v} .+11 \nabla$. THIMMERS, Ceramio. $30,50,70 \mathrm{pF}, 9 \mathrm{~g}$; 100 pF.
$100 \mathrm{pF}, 1 / 3 ; 250 \mathrm{PF}, 1 / 6 ; 500 \mathrm{pF}, 750 \mathrm{pF}, 1 / 8$. RESISTORS. Preferred values. 10 ohmes to 10 meg.

 10 n to 10 meg. Ditto, o \%, 100 a to meg. $1 /$ ${ }^{5}$ WHatt $\}$ Watt $\}$ WIRE-WOUND RESIRTORS 10 Watt $\} 25$ ohens $\rightarrow 10,000$ ohras
12.5 K to 50 K 10 w

| AMERICAN "BRAND FIVE" PLASTIC RECORDING TAPE |  |  |  |
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| Double Play <br> Long Play | 710, reei, 2,400 t | 60--1 | Bpare <br> Plastio <br> Reels |
|  | 5in reel, 1,200ft | $37 / 8$ $351-$ |  |
|  | 7 in. reel, $, 1,8000 \mathrm{th}$ Bilu. reel, 1,2000 | ${ }^{33 / 8}$ |  |
|  | 315. reel. Doult | 18/6 |  |
| Standard | 7in. reei, 1,2001t |  |  |
|  | Stan real, tiout |  |  |

O.P. TRANBYORMERS. Heavy Duty 50 mA, 4/6 M.P. Tature, 884, etc., 5/B. Multiratlo push-pull, $7 / 6$.



|  |  |
| :---: | :---: |
| tappad \& 7.28 . Rectitter 6.3 \%. 1 a. 5 \%. | $22 / 6$ |
| MINLATURE, 200 จ. $20 . \mathrm{maA}, 6.8$ จ. | 10/8 |
| MIDAET, 220 \%. $45 \mathrm{~mA}, 6.8 \mathrm{v}^{2} 2$ 2 | 15/8 |
| 8MALL, $220-0-220.60 \mathrm{~mA}, 0.3$ \%. 3. | 178 |
| STD, $250-0-250,65 \mathrm{~mA}, 6.3$ 7. 3.5 \& | $17 / 8$ |
| HEATER TRANS, 6.3 v. $1 \nmid$ A 7 | $10 / 8$ 818 |
| Ditto 1.4, 2. 3. fi: 5. 6.3 F. 1 1 A | $8 / 6$ $82 / 6$ |
| Ditto Tapped. ${ }^{\text {a }}$ to 30 v., 24 | 22/6 |
| AUTO TRANs. $110-240$ v. 150 | 82/6 |

ALADDIN FORMERS and core, tin., 8d. : fin. 10 d alin. Formers
 MALNS DROPPER. 3 in . I 13 in . Adj. Bilders, MALNS DROPPER. $4 / \mathrm{in}$. 2,4 gmp., 1000 ohms, $4 / 3$. LINE CO'RD. 0.3 amp. 60 ohms per ft, 0.9 acap., 100 ohme per fi, $2 \cdot$ way, $1 /$, , per ft, 3-way, $1 /-$ per fi. LOODBPEAKER P.M. 8 ORM, Sin. Rola, 17/日,
 Rin. Tweeter, $25 \%$. 121 I . R,A., $30 /-14 \times 8$ in. $45 /-$ STENTORIAN HFIO12. 10in. 3 to 15 ohme 10 w., $96 /$ 12in. Bairer 15 watt 8 ohms or 15 ohiss, $90 /-$ 12in. BAKER FOAM SUSPENSION. 15 Ohm, 48 CRISTAL DIODE. G.E.C. $2 /-0$ GEX34. $4 /-$ RIGH RESISTANOE PHONES. ${ }^{4,000} 7 / 6$ pr.
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| 1Rs | 7/616E8G | 7/6 | EABCS | 8/8 | HVR2A | 8/8 |
| 1 180 | 7/618L6G | 10/8 | \#B91 | 61. | MU14 | 9/8 |
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ALUMINIUM CHASSIS. 18 s.w.g. undriled. With 4 sides, riveted corbers sind isthice 0 ying holes, 2tio. sldes, 7 I 41 m ., $4 / 6 ; 9$ x $110 ., 10 / 8 ;$
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CONDENSERS. 1000 上F, 7 kV. T.C.C., B/6, 500 or $1,000 \mathrm{pF}, 30 \mathrm{kV}$., $9 / 8_{4} 0.1 \mathrm{mifd} .7 \mathrm{kV} ., 9 / \mathrm{B}$, Tubular 500 v, 0.001 to 0.05 mefd. $\theta d .0 .1,1 /-1$ $0.25,1 / 8 ; 0.5 / 500$ ₹., $1 / 8 ; 0.1 / 350$ V., $0 \mathrm{~d} . ; 0.1 / 1000 \nabla_{0}$ $1 / 8 ; 0.01 / 2,000$ V., $1 / 9 ; 0.1$ rafd.; 2,500 volk, $3 / 6$, CERAMIC CONDS $300 \mathrm{~F}, 0.3$ yf to 0.01 told., 8 Cd SILVER MICA CONDENSERS, $10 \%$ b 5 W 500 $\mathrm{pF}, 1 /-600 \mathrm{pF}$ to $3,000 \mathrm{pF}, 1 / \mathrm{s}$. Close tolerance $( \pm 1 \mathrm{pF}) 2 \mathrm{\mu F}$ to $47 \mathrm{pF} .1 / 8$, Ditto $1 \% 50 \mathrm{pP}$ to $816 \mathrm{pF}, 1 / \mathrm{\theta} ; 1,000 \mathrm{pF}$ to $5,000 \mathrm{pF}, 2 /$.
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\begin{aligned}
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& \text { coll, serial coll Osclletar coll, two I. P. trana } \\
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& \text { Clrcuit book using four } 6 \text { AM } 6,2 / 8 \text {, } \\
& \text { COMPLETE JABON F.M. KIT FMT1. } \\
& \text { with get of } 4 \text { Falves, 26.6.O. Letails B.A.E. }
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FULL WAVE BREDGE SELENIUM RECTIFIRRS.
 250 F. for oharging at 2 , is of $12 \nabla_{\mathrm{g},} 11$ ampa, 16/6: 260 . 2 . $17 / 6$, 4 mapa., $82 / 6$. Cireut lncluded YALVE DATA, Vol, $1,2, \$$ or $4,5 / \mathrm{e}$, each. TOGGLESWITOHEB, B.P. 8/- D.P, 8/6, D.P.D. $4 /-$ WAVECEANGE SWITCHRS
5 p. 4-wave 2 waier-long plone
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instruction 1？d．Total cost of all mecessary partorif Total cost of all necessary parts oni $49 / 9$.
PRACTLCAL WIRERLSS SUPER NO POCKETPURTABLEMADIODESIGN Dlagrams etc．． $1 / 9$ or All parts including printed circult and first grade transistors 48．19．6．
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ER．LiUVI＇SMuUTHING CHUNES， $200 \mathrm{mA.}.{ }^{3}-5 \mathrm{H} .00$ chms．Parmeko 8／b； 100 EnA． 5 H． 100 ohrus $2 / 11$ ； 150 mA. 10 H ． 50 ohms $8 / 8 ; 80 \mathrm{~mA}, 20 \mathrm{H}$ ． 900 ohms $5 / 8 ; 120 \mathrm{~mA} ., 12 \mathrm{H} .100$ ohrms 8／9：50 mA． ohms 8ig； 60 mA ． $5 / 10$ H 100 mA ． 10 H ．idu ohms 8／8； 60 mA．，5－10 H．， 250 ohms E／11． RE．GUVT．MAINS THANSFORMETS
 $250-0-20$ ． 60 m．a． $6.3 v .2 \mathrm{a}$ ．
－ $11 / 9$ $250-0-20 \mathrm{v} .60 \mathrm{~m} . \mathrm{a} .6 .3 \mathrm{~V} .2 \mathrm{a} . \quad \because \quad \because \quad 18 / \mathrm{B}$ $270-0-275 v .100 \mathrm{~m} . \mathrm{a}, 6.3 \mathrm{v}, 7 \mathrm{a} .5 \mathrm{v} .38 .1 \mathrm{M}$ Q1／8


 5V．10a．Parmeko
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$450-0-450 \mathrm{v} .100 \mathrm{~m} . \mathrm{a} .6 .3 \mathrm{v} .1 .5 \mathrm{a} .4 \mathrm{v}, 3.5 \mathrm{a}_{\mathrm{a}}$.
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\＆VOLI ACCUMULATORS
Varleys small size $4 \times 34 \times 1+1 \mathrm{n} .2 \mathrm{~F} .14$ A．H．brand new 日／8 ea． 3 tor $15 /$ en $^{2}$ ． 17 FIELI TELERHONES

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## A crystal

 frequency marker

FIg. 1.-The circuit of the unit.

Crystal frequency markers are intended to give frequency check points over a wide range, so that receivers or signal generators may be accurucly calibrated. The marker described here has a uscful coverage of $100 \mathrm{kc} / \mathrm{s}(3,000 \mathrm{~m})$ up to about $15 \mathrm{Mc} / \mathrm{s}$ ( 20 m ) or more, the maximum frequency depending on the sensitivity of the receiver.

## The Circuit

The circuit is shown in Fig. 1, and any high slope, R.F. type pentode may be used instead of the 6AM6. With some valves pin connections will of course be different, while octal valves will require an octal holder instead of the miniature holder. Two crystals are provided for, but it is perfectly in order to use a single crystal only, as will be explained. The current required by the valve may usually be drawn from the receiver. The marker may be permanently connected-the heater switch being opened when it is not required. In addition to the heater switch, the 3 -way rotary switch, which allows cither crystal to be sclected, has a central "off" position which interrupts the H.T. circuit. This allows the marker to be turned off and on again without waiting for the heater to reach working temperature. It also allows the marker signal to be interrupted at any time. so that it can be casily distinguished from other carriers which may be heard on crowded S.W. bands.

## Operating Details

Constructors who have not used a crystal marker will probably find details of the method of operation useful. When the $100 \mathrm{ke} / \mathrm{s}$ crystal is in circuit. the output from the marker is at $100 \mathrm{kc} / \mathrm{s}$ and multiples of $100 \mathrm{kc} / \mathrm{s}$. As example, its signal will be heard at $200 \mathrm{kc} / \mathrm{s}(1500 \mathrm{~m}), 300 \mathrm{kc} / \mathrm{s}(1000 \mathrm{~m})$ and so on. In terms of wavelength, the marker signals are widely spaced on long waves, but much closer together on medium waves, and closer still on short waves. As $100 \mathrm{kc} / \mathrm{s}$ is the same as $0.1 \mathrm{Mc} / \mathrm{s}$, the signal will be heard at $0.1 \mathrm{Mc} / \mathrm{s}$ intervals troughout M.W. and S.W. bands. For example, on the M.W. band of about $600 \mathrm{kc} / \mathrm{s}$ to $1500 \mathrm{kc} / \mathrm{s}$, the marker signal will be present at $600,700,800$, 900 and $1000 \mathrm{kc} / \mathrm{s}$, and $1.1 \mathrm{Mc} / \mathrm{s}(1,100 \mathrm{kc} / \mathrm{s})$, etc., up to $1.5 \mathrm{Mc} / \mathrm{s}(1.500 \mathrm{kc} / \mathrm{s})$. The same result is obtained on S.W. bands. Harmonics hecome weaker, and this sets the upper limit at which the marker signal can be heard. With a highly sensitive recciver, harmonics may be heard up to $30 \mathrm{Mc} / \mathrm{s}$ ( 10 m ).

## Receiver Calibration

For receiver calibration, the appropriate marker harmonics are tuned in, and the receiver scale is drawn up to agree, or the dial readings are noted down. For example, marker signals at $3 \cdot 5,3 \cdot 6,3 \cdot 7$ and $3.8 \mathrm{Mc} / \mathrm{s}$ will accurately set the limits of the " 80 m " amateur band. Medium wave tuning scales, or commercial S.W. bands, can be calibrated in the same way.
In the unit shown, crystal 2 was a $1750 \mathrm{kc} / \mathrm{s}$ crystal. When in circuit, this gives marker signals at $3.5 \mathrm{Mc} / \mathrm{s}, 7 \mathrm{Mc} / \mathrm{s}$, etc., in addition to the fundamental ( $1750 \mathrm{kc} / \mathrm{s}$ ), for easy identification of
amateur bands, with any receiver. Almost any crystal which may be to hand can be used for a somewhat similar purpose. It is also quite usual to have a $1 \mathrm{Mc} / \mathrm{s}$ crystal, to give tuning points at $1 \mathrm{Mc} / \mathrm{s}$ interyals, the $100 \mathrm{kc} / \mathrm{s}$ crystal then being used to fill in the $0.1 \mathrm{Mc} / \mathrm{s}$ tuning points between those obtained with the $1 \mathrm{Mc} / \mathrm{s}$ crystal.


Fig. 2 (above).-Above chassis view of the marker.

It is perfectly in order to omit the second crystal and change-over switch. This will not in any way influence the accuracy of calibration at $100 \mathrm{kc} / \mathrm{s}$ intervals, but will mean that the $100 \mathrm{kc} / \mathrm{s}$ marker signals will have to be counted off fairly carefully; or known stations or bands will have to be tuned in, to identify the particular harmonic.

## Unit Construction

Dimensions and layout are of little importance, but a small chassis about $3 \frac{1}{2} \mathrm{in}$. x $4 \frac{1}{2} \mathrm{in}$. x 2 in . deep will be convenient. Fig. 2 shows a suitable layout. The 100 pF variable condenser, which is used for exact frequency setting of the $100 \mathrm{kc} / \mathrm{s}$ crystal, must be insulated from the chassis. This can be done by having a clearance hole in the bracket, and using insulating washers. Some surplus condensers have mounting bushes isolated from the moving plates-if this is the case, insulating washers are

unnecessary. The spindle is slotted to receive a screwdriver blade.
Fig. 3 shows underchassis wiring, the 3 -way switch actually being mounted on the chassis runner. An octal holder is shown, for an octal based $100 \mathrm{kc} / \mathrm{s}$ crystal, but the holder should of course suit the crystal to be used. The second holder was for surplus 3 in. spacing crystals.
A small tagboard provides anchorage for power supply leads, and some of the components. Power requirements are $6.3 \mathrm{~V}, 0.3 \mathrm{~A}$ for heater, and approximately 5 mA at 200 V to 300 V for H.T. Wiring should be reasonably short and direct, and well insulated.
If the second crystal is not to be used, wire the $100 \mathrm{kc} / \mathrm{s}$ crystal holder directly to the 100 pF condenser fixed plates, and tag 1 of the B7G valveholder. A second toggle switch in the H.T. circuit will then allow the marker to be switched off, with heater running.

If a separate power pack is required, this can use a small converter or eliminator transformer, and metal rectifier. Two $8 \mu \mathrm{~F}$ or similar smoothing condensers will be adequate, and the usual smoothing choke may be omitted, a 2 k resistor being employed instead. In addition to about 250 V for H.T., the transformer must have a 6.3 V heater secondary.

When drawing power from the receiver, any convenient means of connecting the leads may be employed. It may be feasible to mount a socket on the receiver chassis, and to fit the marker leads with a plug to suit.


The Prototype Marker Unit

## Frequency, Adjustment

Crystal No. 2 is used for quick location of bands, with an entirely uncalibrated receiver, as described, and no attempt is made to adjust its frequency, as the $100 \mathrm{kc} / \mathrm{s}$ crystal is used for actual, exact calibration. With the $100 \mathrm{kc} / \mathrm{s}$ crystal, the 100 pF pre-set condenser permits a very small shift in (Continued on page 437)

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(Contimued from page 434) frequency. The exact setting of this condenser catn be found by beating the mather against the BBC $200 \mathrm{kc} / \mathrm{s}$ ( 1500 m ) Light Programme, or against the National Physical Laboratory signal radiated on $2.5 \mathrm{Mc} / \mathrm{s}$. By this means. an extremely high degree of accuracy indeed can be reached.

If a new $100 \mathrm{ke} / \mathrm{s}$ crystal is purchased and used with the parallel capacity scting specified by the maker, the standard of accuracy will be high enough for all practical purposes. But it is quite easy to check against one of the standards mentioned. so this is in any casc worth while.

To set the $100 \mathrm{kc} / \mathrm{s}$ crystal agains: the $B B C$ 200ke/s transmitter the frequency


Fig. 3.-The wiring diagram of the crystal marker. of which is maintained to high standards) it is neeessary to arrange that the Light Programme signal and marher signal are of roughly the same strength. This can be done by removing the receiver aerial, or using a short piece of wire for aerial. and also ig modifying the coupling between marker output lead and receiver acrial lead. An exact balance between signal strengths is in no way necessary, but adjustment will be difficult if one signal almost completely swamps the other.
With the receiver tuned to the Light Programme, and the marker working, a very low-pitched audio tone. or flutter, may be
round a short insulated wire inserted in the receiver acrial socket. The receiver aerial may then be temporarily detached.

## Tuning Indication

If the receiver has a tuning eye or meter, this will show fluctuations when the marker signal is tuned in. The marker signal is unmodulated. That is, it consists of a radio lrequency carrier only, with no audible tone. With TRF receivers, the signals can be located by advancing reaction until the detector is just oscillating. The marker harmonics heard. This is the difference between erystal harmonic and BBC frequencies. and the 100 pF trimmer is adjusted with an insulated blate so that the tone or flutter falls in frequency, and ccases.

If the receiver has a tuning eye or buning meter, this will show fluctuations in sigoal strength when the marker tuning is almost exactly correct (say within $1 \mathrm{c} / \mathrm{s}$ ). If the ec is no cye or meter, a rise or fall in background noise or volume will he apparent. A higher standard of accuracy would be quite poinless for all normal purposes.

## Receiver Calibration

The method of using harmonics has been described. Through the L.W. and M.W. bands, the marker signals will be very strong, and only loose coupling between marker output lead and receiver acrial input will be necessary.

For higher frequencies, the marker signal will be weaker, and tighter coupling will be required. It will be necessary to bring the marker output lead near the recciver acrial lead, or to twist it


An under-chossis view of the unit.
can then be tuned in at $100 \mathrm{kc} / \mathrm{s}$ intervals, in the same way as broadcasting stations. A receiver with BFO (beat frequency oscillator) will make the harmonics audible in a similar manner.

receiver which weighs only 6lb 13 oz . The price of the "Courier" is 25 guineas and it is made by Pam Radio and Television Ltd., 295 Regent Street. London, W.l.

## TRANSISTOR RADIO FOR STANDBY NAVIGATION

A PORTABLE transistor receiver, called the "Navigator" is now being marketed in this country by the Zenith Radio Corporation. This set, as well as receiving normal broadcasting stations, is suitable for standby navigation purposes by private pilots and yachtsmen, as it covers weather stations and L.F. directional beacons.
The "Navigator", which covers the 550 $1600 \mathrm{kc} / \mathrm{s}$ and $150-400 \mathrm{kc} / \mathrm{s}$ bands, is an 8 transistor set with 500 mW output. It weighs only 5 lb .

The "Navigator" is fitted with an accurate and easily read azimuth scale for taking cross-bearings and has precision vernier tuning on the main control. The price of this receiver in this country is f69 6s. IId. The agents for the Zenith Radio

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# A Sensitive Multimeter 

## A $10,000 \Omega / \mathrm{V}$ INSTRUMENT

(Continued from page 296 of the August issue)


By R. Murray-Shelley the August issue) should only be used as a guide, the exact lengths being found by a process of trial and error, using another meter of known accuracy in the circuits of Fig. 2. The additional components required for the construction of the shunts are, a 3 V dry battery, a 1 k potentiometer (pre-set, linear), a $2 \cdot 2 \mathrm{k}$ resistor and a $50 \Omega$ resistor.

 vary the length of wire in the shunt until the readings of the two meters are the same for a given current (e.g. for the 10 mA range, the multimeter reads full scale when the check meter reads 10 mA ).

Somewhat lower accuracy can be tolerated for the 1 A range, and thus the shunt should be adjusted so that while the check meter reads 100 mA , the reading of the multimeter is 10 , assuming that the meter is calibrated from 0 to 100 , i.e. 0.1 of the full scale deflection. The reason for this is that a nomal dry cell battery cannot supply 1 A for cven a short period and thus calibra-
tion must be achieved using a Jower current.
When the correct lengths of wire have been obtained, they should be wound either on a small choke former or a 2 or 3 W resistor of at least 5.000 ? resistance. The purpose of this is merely to act as a carrier for the wirc. Resistance wire is occasionally difficult to obtain. and in the prototype, wire removed from a $50 \Omega 5 \mathrm{~W}$ resistor was used very satisfactorily. Insulated copper wire ( $40 \mathrm{~s} . \mathrm{w} . \mathrm{g}$.) was used for the 100 mA shunt, and $28 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. for the 1 A shint. The formula for calculating shunt values is:

$$
R=\frac{I m \times R m}{I-I m}
$$

where I is the maximum current which the meter is to measure in amps, and Im and Rm are as in the August issue.
The shunt for the 1 mA range has a value of over $100 \Omega$. This is most casily made by using a 1002 close tolerance high-stabilty resistor in series with a wire-wound resistor made as described above.


Fig. 2.-These circuits illustrate the use of a meter of known accuracy to calibrate the multimeter under construction.

## Tolerance

The resistors used for the multipliers should be high-stability components of 1 per cent or 2 per cent tolerance. The exact calculated values are not always ohtainable, since only preferred values are usually stocked. in the component list, therefore, when a resistor is not a preferred value, a pair of preferred resistances has been suggested. which when connected in series or parallel, depending on
the resistance in question, will give the required value of the multiplier.

## Resistonce

This instrument has provision for measuring resistances to 1 M in two ranges. The circuit used is quite conventional, and although it is by no means the most accurate method which may be employed, nevertheless, it is adequate for general purposes.

The principle of this method depends on measuring the current flowing in a circuit, the potential across which is known, and thus obtaining the resistance of the circuit from C.am's Law. If the unknown resistance is a part of that circui., then, assuming the resistance of the rest of the circuit is known, the unknown resistance may be found by subtraction. The basic circuit measuring to 1 M is shown in Fig. 3.


Fig. 3.-The basic ohms circuit, measuring up to IM.

## Calibration

In use the meter terminals are first shorted together, and the variable resistor is adjusted until the meter reads full scale. The unknown resistance is then connected across the meter terminals, when the reading of the meter will drop. The new reading is noted and the resistance is calculated using formula:

$$
R x=\frac{R(1-D)}{D}
$$

where D is the fraction of the full scale current, and $R$ is the resistance of the rest of the circuit (in this case 15 k ). This method of calculating the resistance can be rather tedious, and the most accurate alternative method is to draw a graph between resistance and meter readings for various resistances. (Two graphs will be required, one for each range.). Another method of achieving the same result is to add a resistance scale to the meter itself, though this is not so accurate, since the current does not vary directly as the unknown resistance.
It is a good idea to wire the resistance ranges before wiring the voltage ranges, since, then, the precision resistors which are to be used for the multipliers can be wed to catibrate these ranges.
The resistance R14 is used as a range multiplier to achieve a resistance range of up to 10 k . The value of this resistance is given by the formula:

$$
\mathrm{R} 14=\frac{\mathrm{R}}{(\mathrm{~N}-1)}
$$

where $R$, as before, is the resistance of the meter circuit, and N is the ratio of the resistance ranges. In this case:

$$
\mathrm{N}=\frac{1 \mathrm{M}}{10 \mathrm{k}}=100 .
$$

For meters not having the specified characteristics, the value of $R$ is given by (battery E.M.F. $\div \mathrm{Im}$ ) where the battery E.M.F. is the voltage of the cell used-in this case a 1.5 V penlight cell, or a larger U10 cell is recummended. 1 m is the meter full-scale deflection as before (in amps).


The resistance ranges are also useful for testing continuity, at the same time giving some idea of the resistance of the circuit under test.

## Construction

The prototype instrument was housec in a metal box, size $7 \frac{1}{2} \mathrm{in}$. $x 4 \frac{1}{2} \mathrm{in}$. $x 2$ in. deep. This box was of very stout construction, being diecast, and made of zinc alloy. Ranges of meter boxes are available, and the exact size used will depend entirely upon the size of the basic meter movement. There is no real objection to the use of a steel box since the meter used is of the moving coil type, and is not easily affected by external magnetic fields as a moving iron meter might be.
The layout is in no way critical. In the original instrument the multipliers and shunts were carried on a group board attached to the rear of the meter movement.

Connection should be made with stout (16 to 20 s.w.g.) tinned copper wire, insulated where necessary with insulating sleeving. The rectifier elements can be wired directly to the range switch and to the A.C./D.C. switch. The range switch has three wafers. each being 1 -pole, 12-way, giving a composite 3 -pole. 12-way switch. If this is difficule to obtain, a 3-pole, 11-way switch could be used
(Continued on page 445)

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## (Continued from page 442)

in its place, though one, or possibly two ranges would now have to be omitted (e.g. 1.000 volts A.C. and D.C., or the 1A range of the 10 k range). The purpose of the third wafer of the range switch, S2c, is to avoid any possibility of damaging the meter movement on the higher current ranges. This might well happen if the positive terminal of the instrument were to be connected directly to the pole of the wafer, S2b, and a faulty contact were to develop on S2b. The result would then be that the shunt would be removed from the circuit, and the full current in the circuit under test would be applied to the meter movement, in most cases causing it to burn out.
It is important when wiring the range switch to note carefully which position of the switch corresponds to which range.

All connections to the multimeter are made using only two terminals, and these terminals should be of the type which are totally insulated from the panel. The cell is conveniently held with a Terry clip. Connections to the cell may be soldered, since its life is many months.

## Resistor Ratings

The resistors used for the multipliers need only be of $\frac{1}{2} W$ rating, since they will only be called upon to carry a very small current. In the case of an instrument using a 1 mA meter, however, the resistors should be rated at 1 W , at least for the 300 V and $1,000 \mathrm{~V}$ ranges.
The power dissipated by a resistor is given by the formula: Power (watts) $=(\mathrm{Ir})^{2} \times \mathrm{R}$,


Fig, 4.-The layout of controls on the front panel.


The underchassis view with the wiring complete.
where Ir is the current in the resistor in amps, and R is its resistance.

The current in the resistor will be the same as that in the meter, and thus the maximum current which a multiplier will be called upon to carry is equal to the full scale deflection of the meter. Using this relationship the ratings of the resistors required for any meter movement can be calculated.

## Soldering

It is particularly important when constructing equipment of this kind that all soldered joints should be of low resistance, and not of the "dry"" type. Cored solder should always be used, and the iron should be clean, and sufficiently hot that the solder flows easily. The components to be soldered should be cleaned carefully, particularly the switches, if these are of the "surplus" type. In this case, the switch contacts themselves should be cleaned, first by scraping them, and then by using a proprietary switch cleaner, preferably one which contains switch lubricant.

## Taking Readings

When using the completed instrument, always start with a high range, working down to the lower current and voltage ranges. This avoids any possibility of damaging the meter movement. The instrument should be disconnected from the apparatus when changing ranges. The test leads can be about 3 ft long, and they should terminate in a pair of test prods, if possible of the retractable type.
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