## PRACHCAL 16 WWIRETESS




HI－FI IIEADPIIONES
 FI phones use high qualtty permanent magnetic speaters with regu－ coll．The softrubber ear moulds give correct spacing for optimum acoustic load．each Fi transformer to ensure the finest music and voice reproduc－ tion．Supplied free is a small transformer unit which steps impedance up to 4000 ohms． Only 15／－P．\＆P． $2 / 6$.
COMMAND TRANSMITTELRS Complete with ald valves，crystal ${ }^{\mathrm{etc} .} 3-4 \mathrm{Mc} / \mathrm{s} .5 / 5 /-4.5 .3 \mathrm{Mc} / 6.45 /-$ COMMAND RECEIVERS $6-9.1 \mathrm{Mc} / \mathrm{s}$ ． $60 /$－P．\＆P． $3 / 6$ ea． CRISTAI．MIKE INSERTS 11n．dia． $70-6000 \mathrm{c} / \mathrm{s}$ response．Brand new． 2 for 12 －．
－PORTABELETTRANS／ －FECEIVER No．18．A self－ contalned Trans／Recelver for approx． 10 miles．Freq． $6-9 \mathrm{Mc} / \mathrm{s}$ ． （50－3R．3 metres）．Valve line－up： － 3 ARP－ 12.2 AR－8． 1 ATP4．Com－ plete with aerial．H．T．and L．T． 120 meter and all accessories，Welght 20 lbs．Size 8 x 10 玉 17 in ．Only 80／－．Carr．10／－．


These masnificent Instruments will enable you to recelve maximum signal strength on all S．W．receivers． Precision cali－ brated control 12／8．P．\＆P． $2 / 6$.

MICROPIONES．BRAND NFW．Throat magnetic，4／6； Throat carbon 3／6：No． 8 carbon with switch 6／6；No． 7 moving coll 7／8．Tannoy power mikes only 5／＝，Crystal mike 22／8．

## PORTABLE RADIOPHONES MODEL MK II

## Brand New British Army Portable Transmitter Receivers．

Destgned for rellable vatce intercommunication operating up to 10 miles depending upon obstructions end elevation．The combined Transmitter Receiver covers the whole frequency range between $74-9 \mathrm{Mc} / \mathrm{s}$ ．and is fully tunable on both Transmitter and Recelver．Simple and a delight to operate as ell controls are mounted on the front panel of the set and clearls marked．Operates from standard dry battertes 3 v．L．＇T．and 120 v．H．T．Incorporates 5 vaives：R．E．Amplitier．I．F．Amplifer．Second
All sets are supplled complete with all accessorles comprising dynamic sound powered headphones， electro magnet supersensitive microphone，4ft．aerial．junction tiox，battery connection decauls and fall circuit diagram．

## PIRICES：

60／－EACHI（P．\＆P．4／－）。


NEW：TRANSISTOR RADIO With Miniature Sbraker． Simple instructions enable radio Gives reception over the entire broadcast band Fach kit is supplicd with illi latest mindature parts including： $\star$ two transistors $A$ ferrite rod $\star$ speaker $t$ coloured plastic case t step bs step H1ustrated
 Battery $1 /$ ex

VHF IRECEIVEIR 1392
15 valve superhet covering $5-150 \mathrm{Mc} / \mathrm{s}$ which includes eur bands．Complete with sow motion tuning tuning meter．circuit diagram． etc．Here is your appor－ tunity to explore the VHF bands on a realls sensitive recelver offered at a frac－ ton of original cost．


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This basic scope represents one of the flnest buys we have ever made Contains Brilliance．Focus．Gain and 2 speed time base controls． eparate $x$ piace terminals．Signa Generator modulated at 2 trequen－ cies over with 12 valves．VR b9A tube complete clrcult and technlcal comple

ONLY 97／6 carr．12／8

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A really sensitive dynamic earphone of exceptionally fine quality．Pro－ vides clear reproduction of music as well as speech．Fully Guaranteed and 3 feet cord sub－minlature plug and socket．

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| $8 /-\frac{\text { Each }}{\text { Post } 1 /-}$ |



SIGNAL GENERATOR
Morlel SWO－300 150 ke／s－ 300 Me／s．Frequency Range： 150 $\mathrm{kc} / \mathrm{s}-150 \mathrm{Mc} / \mathrm{s}$ on tundamentals
bands $), 150 \mathrm{Mc} / \mathrm{s}-300 \mathrm{Mc} / \mathrm{s}$ or 6 bands）． $150 \mathrm{Mc} / \mathrm{s}-300 \mathrm{Mc} / \mathrm{s}$ or harmonics．Calibration Accuracy Within 1 per cent，Modulatios． Internal and external．Attenus－ ation：To－40db．Output：Faclli－ ties for high and low．Power $7 \times 10 \times$ 5in．Complete with test and instruction manual ONLY £14．19．6．Carriage 5／6．


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POCKET VOLT TEST MFTEIS O－250 D．C．ranges： 25 v Completo with test prods and leather prods gnd eather case． quantity Only 12／6．P．\＆P．1／6


R．C．A．COMMUNICATION RECEIVEIR AR－88 I．E．Hange： $73-550 \mathrm{kc} / \mathrm{s}$ and $1.48-30.5 \mathrm{Mc} / \mathrm{s}$ ． Undoubtedly the finest com－ duced for service and laboratory duced for service and laboratory chase enables you exciusive pur－ Rolls－royce of communt enion recefvers at the communication offered．Fully guaranteed．ONLY £37．10．0．carr． $50 /$－．

ACCUMULA－ A．H．（unspillable） Ideal for 6 and 12 Brand new．Original cartons．Size 4 da ．$x$ P．\＆ $\mathrm{P}, 1 / 6,3$ for $15 / \mathrm{F}$ ． P．\＆P．3／6， 6 for



WIRELESS SET No． 19 MK． 11
Incorporates TX／RX covering 2－8 Mc／s（ $37.5-150$ metres），and inter－ com．amplifier．Completa with 15 valves， 500 microamp check and tuning meter，circuit and instruction book．ONLY 65\％． \％Carr．10／

T．C．S．RECEIVEIR． 1.5 to 12 Mc／s 7 －valve superhet．bullt like a dream．Panel controls： H．F．Gain．A．F．Gain，C．W． Pitch，band－switch，mod．C．W．
Switch．power switch giound switch．power switch，ground and aerial posts，M．O．or crystal frequency switch，speaker jack， Card nolder to log 30 stations， hand vernier tuning knob turn－ plate behind hafrlined window． ptate behind hair lined window． anti－backlash gears used． $90 \%$
complete．store solled．offered


Exciting New Product Exclusive to RELDA

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MODEL FC－8 Induction Pick up coll enabiing conversations to be pleked up without tapping of wires or special telephone circults．No electribal con－ nections！Simply place telephone on the pick－up platform and sonnect lead to the input of any medium gain lectrier or direct to any tape disc．or wire recorder．The coll th lectrostatically shielded to minimise hum pick－up．Brand new complete with 5it．shielded catie．FC－8 orily．18／－plus $1 / 6$ p．\＆D．

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This amplifier as illustrated, made by a leading manufacturer. Mullard valyesECC83, EL84 x EL84, EZ80. Bass Treble and Volume on remote panel. Elegant Knobs. OUR PRICE one month only £4.16.6 plus P. \& P. 3/6.


HARVERSON SUPERHET 4 KIT

A medium and 10ngwave superher, incorporating two I.F. stages modern B9 valves (UCH8I, UBF89, UCL83, U785), built-in ferrite rod aerial. All you need supplied from theoretical wiring diagram to last nut and bolt (main components ready mounted), including an attractive contemporary styled cream plastic cabinet with gold trimmings. Size IIf $\times \frac{1}{1}$ $\times 6 \frac{1}{2} \mathrm{in}$.

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500-500 Twin gang conderssers with geared slow motion drive. $3 / 6$ ea. 36/-per doz. P. \& P. 6d.
WIRE WOUND POTS
12 Wire wound Calvern Pots $\begin{array}{ll}\text {-all different values } & 10 / 6 \\ \text { P.\& P. 9d. }\end{array}$

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$\begin{array}{ccccc}\text { G.C.Z. } & \cdots & \cdots & \cdots & 1616\end{array}$

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LATEST B.S.R. MONARDECK (single speed) $3 \frac{1}{2} \mathrm{in}$. per sec., simple control, uses $5 \frac{3}{5} \mathrm{in}$. spools ........................................ Ł7. 5. 0 plus $5 / 6$ carriage and insurance (tapes extra). TRUVOX MARK III TAPE DECK. New and Boxed ...........................................€10.6. 6 Plus 6/- carr. and ins. (tapes extra).

## STEREOPHONIC AMPLIFIER £5.10.0 Complete with 2 Loudspeakers <br> Plus 4/6 P. \& P.

This is a compact amplifier embodying the latest features and giving a high standard of reproduction, with ample volume. Supplied complete with valves (ECL82, ECL82, EZ80), panel, knobs, etc., and two specially selected 3!) matched loudspeakers. We only have a few, and we will never be able to repeat this offer at such a low price. Don't risk disappointment. Order now.


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Complete and ready for your cabinct, 4 valve superhet chassis, complete with valves, ferrite aerial, dial and chassis, Coniplete line-wp-UCH8I, UBFB9, UCL83, UY85. Long'and Mediurn wave coverage.

PRICE E4.19.6 P. \& P. 3/6. 18/6 each. P. \& P. I/9.

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## TRANSISTOR RECORD PLAYER CASE

A few only-Transistor record player cases in light grey cloth-complete with motor board. Size: $12 \times 8 \times 6 \mathrm{in}$.

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108K 10 -inch. New and boxed, 151-, plus 61- P. \& P.
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To suit the above, 219 each. P. \& P. 3d.

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Super quality heavily woven frat, 54 inches wide. Usual price 50/= per yard.
P. \& P. 1/-

OUR PRICE $12 / 6$ per yard.

## CYLDON 12 CHANNEL TURRET TUNERS

New purchase offered at still lower price. I.F. 33-38 Me/s. Complete with PCC84 and PCFBO valves and 8 sets of Coils for 5 Band I Channels and 8, 9. 10 Band IIf. New and $\begin{aligned} & \text { unused. Value over } £ 7 . \\ & \text { OUR PRICE, post paid }\end{aligned} 32 / 6$

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50 mixed P.F. Condensers and 50 mixed Resistors. An assorement of useful values. All popular sizesall new-a must for the serviceman and constructor. ONLY $10 /-$
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The product of a renowned maker, this stereo amplifier is composed of "ready-built" units, only requiring interconnection. This system has the advantage of being adaptable to fit any cabinet. Each unit is made from first-grade components, and valves used (ECL82, EZ80 range) are genuine Mullard. The comprehensive instructions supplied make the simple interconnection of units easy even for the novice.

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TWO MIDGET AMPLIFIERS each of $3 W$ output, good reproduction trom boch your stereo or monaural records. Both amplifiers complete with well-designed O.P. transformers providing perfect matching 3-7@ speakers, and have remote bass, treble

CONTROL UNIT, is a flying panel with three 2 -gang pots, enabling the bass, treble and volume controls of each amplifier to be conveniently positioned. Supplied with atcractive cream and goid knobs.
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A chassis of distinction, by a famous maker. Covering Long, Med, \& Short Waves, plus gram position, this chassis (Size $15 \frac{1}{2} \times 7 \times 6 \frac{1}{\frac{1}{2}}$ in. high) incorporates the latest circuitry, using fully delayed A.V.C., and negative feedback. Controls: Tone, Vol.- On/OH, W/Change (L.M.S. \& Gram). Tuning. Tapped input 200-250 v. A.C. only. An attractive brown and gold illuminated dial with matching knobs, make this one of the most handsome, in addition to being one of the best performing chassis yet offered. Complete with valves (ECH81, EF89, EBC8I, EL84, EZ81). knobs, output transformer, leads, etc. OUR PRICE ONLY plus $4 / 6$ post \& packing.
£9.19.6

GUARANTEED VALVES PROMPT DESPATCH $\star$ PROMPT DESPATCH $\star$ Post 6d. per valve extra

SEPARATE POWER PACK wich valve rectifier, midget size ( $5^{\prime \prime} \times 2^{\prime \prime} \times 3 \frac{1}{4}^{\prime \prime}$ high.)
ISOLATED MAINS TRANSFOR-
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VOLTAGE SELECTION PANEL. Fitted with the "valve base" type of mains i/p selector and a channel output socket.
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CREAM DOUBLE PUSH BUTTON SWITCH of neat design gives positive on/off switching.

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All with Long Spindle and ORMIB 131-| RM-1 Double-pole Switch. $4 / 6$ each. DRM2B $15 / 6$ RM-2
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tmg. $\frac{1}{2} \mathrm{mg}$. 1 meg .2 meg. $2 \mathrm{RM}-0$ 7/ij $\left\lvert\, \begin{array}{ll}\text { RM-5 }\end{array}\right.$

| 513 | $14 A 86$ | 1716 | $\mid 4 B 130$ |
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| :---: | :--- | :--- |
| $14 I_{-}$ | $\|4 \mathrm{~A}\| 24$ | $28 t_{-}$ | 41. 14 Al 24 28i-

35/= |4RA 1-2-8-3 21/ (FC3I) 6RD 2-2-8-1 121. $\begin{array}{ll}\text { I6RE } 2-1-8-1 & 816 \\ 18 R A & 1-1-8-1 \\ 4 / 6\end{array}$

18RA I-1-16-1 616 (FC|l6)
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JUST OUT. MIDGET SILICON RECTIFIERS. OUTPUT 120 VOLTS AT $\frac{1}{2}$ AMP. TWO IN SERIES GIVE 240 VOLTS

## AT $\frac{1}{2}$ AMP. NO LARGER THAN A RESISTOR. $10 / 6$ EACH.




This set can be built tor $£ 7.19 .6$. Size $6 \times 37 \times 13 \mathrm{in}$. Weight 17 oz . This set covers medium waveband $190 / 500$ metres intermediate frequency. $470 \mathrm{kc} / \mathrm{s}$ using 4 transistors (Ediswan) and 2 diodes on a printed circuit board, plus a $2 \frac{1}{} \mathrm{in}$. moving coil speaker. Instruction book with point to point wiring diagram, 2/6 each. Batteries, PP4 (Ever Ready) 2/- each.

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At last a Meter at the right price for the home constructor. An accurate Audio Power Output Meter. Two ranges 25 mW to 1 W and 1 W to 10 W . Accuracy $5 \%$ matched for $3,15$. 600 ohms with inst. and tech. data. New 3 -in. $0-1 \mathrm{~mA}$ meter.
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## FERGUSON VHF/FM TUNER

This F.M. adaptor is completely self Greatly Reduced contained and can be fitted to any A.C. £13 mains radiogram or radio with P.U.
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P. \& P. 2/-200-250v. A.C. range $80-98 \mathrm{Mc} / \mathrm{s}$. LIST PRICE 18 gns.

## YOUR EXISTING RADIOGRAM OR RECORD

 PLAYER CHANGED TO STEREO BY OUR STEREO ADAPTORTech. Data 2 valves EF80 and EL84. Switch control for Stereo and Mono Dual volume and tone control. Output 3 ohms, suitable for use with Acos GC/71 GP/73 and Garrard GC/10. Complete £2.19.6 with wiring diagram.
P. \& P. 2/6

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Printed circuit technique using Kit of Parts 2 Surface Barrier lype tran- Complete sistors, one A.F. Trans and one Diode. Ferrite aerial, fully tunable on LW and MW. Uses 2 U16 11 $\frac{1}{2}$ volt batteries. And easily assembled by the beginner.


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A must for alt home constructors. Basic movement $300 \mu \mathrm{~A}$. Sensitivits $1,000 / \mathrm{V}$. A.C. and D.C. Ranges $10-50-250-500-$ $1,000 \mathrm{~V}$. D.C. and A.C. $1-100-500 \mathrm{~mA}$. D.C. $2 \mathrm{~K} / 200 \mathrm{~K}$ ohms. Accuracy : $3 \%$ FSD and D.C. $5 \%$ FSD and A.C. $10 \%$ on. Resistance scale. Dimensions $5 \frac{x}{x}$ $3 \frac{1}{4} \mathrm{x} 2 \frac{1}{2} \mathrm{in}$. Complete with test leads, prods and 1.5 V cell.
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FAMOUS MAKE P.V.C. BASE ON PLASTIC SPOOL 1800 ft on 7 in . spool . . 32/6 1200 tt on 7 in . spool . . 21/1200 ft on $5 \frac{3}{3} \mathrm{in}$. spool . . $22 / 6 \quad 850 \mathrm{tt}$ on $5 \frac{3}{4}$ in. spool .. $16 / 6$

## the latest miniature earphone

This wonderful instrument can be used as a baby alarm or inter-speaker system for home or office. One control unit complete with one slave. attractive tygan grill and handsome natural finish

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## INTERCOM UNIT

 wood cabinets. Extra slaves may be obtained.OUR PRICE
£7.19.6
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B.S.R. UA8 comp.
with latest "ful-fi"
cartridge ..
Stereo
$£ 6.19 .6$
Conquest
Garrard RC 120, 4 -speed auto.

9 gns. Collaro Junior. 4Comp. with Arm and P.U.
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This lightweight superbly made instrument enables you to listen in to your radio wherever you are, ideal for all forms of travel, etc. Provides excellent reproduction and can be worn without any discomfort, complete with transparent earpiece, 3 ft . of fine flex, plug and socket.
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SUB-MINIATURE TRANSISTOR PARTS
We can now sunply all sub-miniature parts for the home constructor-transformers. coils, condensers, resistors-from our large stock of component parts.

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Best Quality Record200fting rape

850 ft . on 5 娄in. spool, $18 / 6$.
600ft, on 5 in . spool, 1319.

200ft. on 3 tin. spool,

5/3. COLLARO JUNIOR Single 4-speed Record Player Unit in cream, complete with pick-up fitted turnover cartridge. Special Price $75 \%$. each. Post $3 / 6$

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All above components are brand new and
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Aruente volume control with switch 1016. Ardente wavechange switch, 3 position $3^{\prime 9}$. Celestion $2 \frac{1}{2} \mathrm{in} .3$ ohrms speaker $26 / 8$.
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Send for our Detalled Price List

## T.S.L. Subminiature

 ComponentsSolid Dilectric Tuning Condenser Type PVC-2M Capacity Aerial 196pF OSC 87pF, size $1 \times 1 \times$ $\frac{1}{2}$ in. 1716 each
I.F. Transformers. set of 3 and Oseillator coil, size of each $\frac{7}{10} \mathrm{in} . x \frac{7}{16} \mathrm{in} . x$ $\frac{1}{2}$ in high, pen see $211 /$. Driver Transformer type TR 190/EIP: ratio 2 4 , size $8 \times \frac{1}{2} \times \frac{1}{2}$ in., $6 / 3$ each.
Output Transformer type TR 190/EIE, ratio 5.3 to I Primary Impedance 240 ohms. Secondary 10 ohms. Max Output 800 MIW. size Output 800 NW. size TSL-Lorenz Loudspeakers, $2 \frac{3}{4}$ in. diameter speaker designed for miniature receivers, 4/500 mW, impedance 10 ohms. Freq. response $120-14,500 \mathrm{c} / \mathrm{s}$., price $25 \%$ Elstone Multi-Ratio Output Transformer, 10 watts. 23/.
Westinghouse LW9 Rectifier, $250 \mathrm{v}, 250 \mathrm{~mA}, 19 / 6$. WB Stentorian HF 1012, 10 watts, 95/-ea. BSR Monarch UA8, Rec ord Changer. E6.19.6. Scotch Recording
Tape $150 / 18$, 7 in . spool,
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Mr．T．Belt，London，S．W． 11 ．
＂I have made up one of your 55／－tran－ sistorised radios and I am very pleased with the results
Mr．A．J．Simmonis，Welling．Kent．
＂I purchased from you a wock ago the Pocket 4 Transistor Kit．I put it together last night in $1 f$ hours on switching on the set，il was right on Radio I，uxemburk，I must say thank you，because not only has the set a very attractive appearance． it also behaves fantastically．＊

AF．N．Elliott，Pontynool．
＂I have completed the assembly of your Pocket 4 radio and am pleased to say that it works from the first switching on＂．
Mr．F．Jackson．Ickenham，Middx．
＂I have built the Pocket 4 and am more than pleased with the results＂
Mr．（i．Bamford，is amspate．
＇I find this set even better than you claim it to be and most certainly up to your usual standard of quality．I feel that nobody could fail to build it and get results．Even the first－time－evar novice． as vour circuit diagrams and instructions ara so clear and precise＂．

ORIGINALS MAY BE SEEN AT THIS OFFICE Mr．R．Relt，Newrastle－on－Tyne． ＂I have built your Pocket 5 Transistor set 1 am very pleased wath $1 t^{\prime \prime}$ ．
Mr．IR．Norge，Hirchington－nn－Sea．
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Variometers for W/S No. 19. Fully tested and working, 12/6. P. \& P. $2 / 6$. Filament Transformers. Primary $0-190-210-230-250$ v. 50 C . 5 ec .12 .5 v C.T. at 10 amps., $22.5 \mathrm{v.C.T}$, at 10 amps .; $310.5 \mathrm{v} . \mathrm{C} . \mathrm{T}$. at 1 I amps., $3,000 \mathrm{v}$. insulation. Price $\mathbf{E 2 . 1 9 . 0}$. P. \& P. 5/-. Primary $0-190-210-230-250$ v., 50 C Sec. $110 \mathrm{v} . \mathrm{C} . \mathrm{T}$. at 4.5 amps .; 210 v . C.T. at 4.5 amps., $4,000 \mathrm{v}$. insulation \&1.16.0. P. \& P. 5/\%. Primary 230 v $50 / 60$ c. 67 v. amp. Sec. 16.3 v. $1-6$ amps.; 263 v. C.T. 3 amps.; 36.3 v C.T. 3 amps.; 46.3 v. C.T. 3 amps. 61.12.0. P, \& P. 5/.

Ferranti Transformer. Oil cooled 20/19.5 K.V.A., 3 phase, 50 cycles. Primary 360/380/400/420/440 v. Secondary 2,700/ $2,900 / 3,100 / 3,300 / 3,500 \quad$ v., $2.1 \quad \mathrm{amp}$. Voltage regulation by simple switches on primary and secondary. Weight 1.150 lbs . Price $\mathbf{E} 125.0 .0$. Carriage at cost.
Driver Transformers. Primary 500 ohms imp. Sec, to match two 805 in push-pull. El.7.6. P. \& P. 5/..
Vibrator Unit. 12 v./l60 v. 35 mA . Exceedingly well filtered and smoothed, excellent for car radios. New, including 6X5G valve and vibrator, $17 / 6$. P. \& P. 5/\%. Moving Coil Round Hand Microphone No. 13. $2 \frac{1}{2} \mathrm{in}$. dia. with press switch, 1216 . P, \& P. $1 /$.
Complete Set of Strong Aerial Rods (American). Screw-in type MP49, 50, 51 52, 53, total length 15 ft . IOin., top diameter 0.615 in ., bottom diameter $0-185 i n$., together with matched aerial base. MP37 with ceramic insulator, ideal for car or roof insulator, $\mathbf{6 2 . 1 0 . 0}$, post free.

PERSPEX FRONTS
（A）on 110：TVG．Heary，Ideai windsureens，ctc．， traction of cost， $\mathbf{1 7 i n} .9 / \theta, 2 l i \mathrm{u} .13 /$－，

## SYSTOFLEX

2．3 min．assorted volours．Twenty－foum 1 yd lengths， $3 / 6$ ．

## 1，000 YDS．ASSAULT CABLE

## P．V．C．eovered stecl，ideas telephone ines，garden

 ing． $8 /=$ drum
## MORSE KEYS

V．G．Anlsh，2／6．Buzzers， $2 / 6$ ．

## INSULATING TAPE

Good quality，balf price， 6 rolls， $2 / 6$ ．

## CERAMIC CONDENSERS

Top quality，rock－bottom prices，3，4，8，8．2．9， 10 $22,15,16,18,20,223.25,30,33,35,47,50,70,82$ $100,110,120,1301150,180,200,800,350,450$ $470,500,600,700,750,800,1000,1500,1800$ $2000,3000,4000,6000$ ．8d each． $5 /$. doz． 12／6 3 doz．

## TUBULAR CONDENSERS

Top muakes． $350 / 500 \vee \mathrm{w} .0 .001,0.002,0.005,0.01$ ， $0.02,0.03,8 \mathrm{~d} ; 0.05,0.1,0.2,0.25,0.5,10 \mathrm{~d}$ ．

## ELECTROLYTIC CONDENSERS

$4 \mathrm{mfd} 500 \mathrm{Fw}, 1 / 6 ; 5 \mathrm{~m} / \mathrm{d} 50 \mathrm{vw}, 1 / 3 ; 8 \mathrm{mfd}-75 \mathrm{v}$ \％ $1 / 3 ; 8 \mathrm{mfd} 450 \mathrm{Fw} .1 / 9 ; 8.8$ midd $450 \mathrm{vw}, 2 / 8$ 16 mfd 350 下w， $1 / 0 ; 15-16 \mathrm{mid} 350+w, 2 / 9 ; 35 \mathrm{mfd}$ 2／9； 50 mfd $12 \mathrm{v}, 1 / \div ; 50-50 \mathrm{mfd} 400 \mathrm{v}, 4 / 6$ ．

FERRITE RODS
6 를 ftr．spprox．， $2 / 8$ ea．； $6 \times 1$ no． $8 / 9$ en．

## BATTERIES

All－Dry，ex＊government， 60 plus， 1.5 v．， $2 / 6$ ea． 6 for 12／6；1 doz． $22 / 6$

## INSTRUMENT KNOBS

White，beautifulig finished．fin．，1／－：lin．，1／3； 1 lin．， $1 / \mathrm{B}$ ．

RADIO AERIALS
Top quality，inaviated，wift．，1／－ex．．，10／8 doz．

## SWITCHES

IAXLEY type， 4 pole， 3 why， $3 / 3 ; 3$ pole， 4 way， $3 / 3$ ； 2 pole， 6 way， $3 / 3$ ； 2 pole， 2 way， $3 / 3$ ；Togzle． cmall． $1 / 9$ ； $5-10$ amp ration，2／3；Spring loaded

## COILS

Techtrad Midget Superhel coll sets，serial and osclilstor．long and medium wave，Polystyrene formers，complete circuit．8／6．

1000 ohm， 10 watt RESISTORS
Ideal smoothing Droppers， $1 / 8$ ea．； $13 /$ doz

## CARBON RESISTORS

$1 / \frac{1}{3}$ matl．3d．ea．： 1 watt and above， 4 d ．ea．；Close tolerance resistors，6d．Uniess shated on order．

## MAINS DRIVEN BELLS

200－230 A．C．，Jncorporates step down traneformer． 9／8 са．

## CHARGING EQUIPMENT

tRANSFORMERS．Primary $0.210 / 240,12 \nabla$ bat terres， 1 amp size， $9 / 9$ ，plus $1 /-$ post． 2 amp sizc， $14 / 8$ ， $1 / 6$ post； 4 amp size， $19 /-1 / 9$ post； 6 aцp size， 25／－2／－post．

VAL，VE HOLDERS
隹保national Ootel．Old Enghen， 7 pin，B3G EA50），3d．ea．，2／0 dos．
Moulded Mazda Octai．3d．es．，2／9 doz．
Internations1 Octal．B7G，B8A，6d．ea．，5／－doz． B7G wilb gkirt，B9A with kkirt，8d，en．，7／－doz crecning Cans，B7G and B9A，6d．ea．．8／－doz．

## FOCUS MAGNETS

Wide angle with foems control．Angle Magnet sype， $9 / 8$ ；Double Magnet type．12／6．Post $1 / \%$ ．

DEFLECTION COILS
thandard，wide angle，low impedance．18／＊，post $1 /$－

## PICK－UP CARTRIDGES

ACOS 22／8 POWER POINT 18／6
LRIG AND REUTER 15／w，SONOTONE 17／6

## Amazing Value！

## STEREO OUTFITS

Constating of two 3 valve（10ば3，LUP1t，UU9） 3－watt malus amplitiers each omplete with Sin．loudspeaker fit weat bakelite cases with stereo changer snd screeted surtio changer snd acreetied \＆11．10．0

## PORTABLE

 RECORD PLAYERSCOLLARO 4 －speed autochanget， 4 wath higb galn amplister．supter two－tone case． 13 gns． complite on bafte with Bkeaker，Ex．10．0．

## TAPE DECKS

B．S．R．＂MONARDECE＂．Latest typee 91．p．s．
 SCRIPTOR．TLIree motora，three speed，TRAN－ 8t i．p．s．，taker 7in．spools，super quality inish，push button controls．$£ 14.10 .0$ ．

## CO－AXIAL CABLE

Scmi－Alr－Hpaced tow loss．1－10 yds，7d．per d．，P．\＆P．1／3．20－30 5ily．6d，pet 5 d ．

## CONNECTING WIRE

 gORIA KIT：Consiating of Tuning Heart， aasembied，with Let I．F．Traneformera．Becond FM．AM and Discrialnator Transortuers， complete kit．Comprehensive instruction bookilet supplied．Booklet only $2 / \%$ KIt，with ECC83 valve， 13.13 .6 ；Kit，less ECC85 valve， £3．5．．

SOLDERING IRONS
50 watts， $45^{\circ}$ pencil tit．16／9．
SPARES FOR OLDER－ TYPE SETS
Framue Output Transformera Blocking Oscllators． Live Output Tranformers．Nean Cons，Focus Coils，etc．：cta（Send S．A．E．for Enquiries）．

## SUPER STEREO KIT

Two Midget Ampllifere 3 F ．Control Unit，Sepurate Power Pack．Voltage selectlon Panci． 3 in． Loudspeaker．Double Push Bntion imitch． Indlcator Light．The KIt 58／8．P．P．3／6．

## JASON TUNER KIT

FMTL ：－E8．15．0．
Lers Power Pack．
PMT3

## UNIVERSAL VOLT OHM MULTIMETER

Reada A．C．D．C．Volts to 1000 ． 3 Rangea at 1000 ohms per voit．D．C．current 3 Ranges to 500 milliamper Resistance Rearling to 200 K in 9 Ranges．Complete with Prods．52／6．P．P．1／6．

BATTERY CHARGERS
6 v ，and 1.2 v ． 1.5 ampus， $49 / 9 ; 6 \mathrm{v}$ ．and 1.2 v .3 amps ，


## POTENTIOMETERS

100 otms $1 \mathrm{ha} . \mathrm{sp}$ ．
200 ohm w．w．in． sp
500 ohm $\mathbf{~ w . w . ~ I n . ~} \mathrm{tp}$
500 ohms w．w．lize spo
100 ohm pres
ik in．sp．
k premet
Ik w．w．jin．ap．
$\because \mathrm{k} / \mathrm{p}$ witch
2k w．w．preset
？2k amail preset
2k 1 1 m ． sp ．
5 k w．w．Itin．ap．
$5 \mathrm{k} \mathrm{w} . \mathrm{w}$ ．presel
10k．lin． Bp ．w．w．
10k $\mathrm{fin} . \mathrm{sp}$ ．
10k premet
$20 \mathrm{k} \mathrm{hm} . \mathrm{sp}$ ．


200 k preset
250 k preset
50 k preset
in．Ohm preset in．obur fin．${ }^{8 \mathrm{p} .}$ ． 1．$\overline{3} \mathrm{~m}$ ．obm 2 m ．obm prese 5 m ．ohm preset
5 in．ohm oniy 25 m ．Ohto $\mp . \mathrm{w}$ ． 22k．w．w．preset $25 \mathrm{k} 1 \nmid \mathrm{ln}$ ．вp． 35k preset ${ }_{25}^{25 k}$ Ha． sp ． 25 t lin． mp ． 25k preset smalh 50k 110．presete 50 k 1 in ．ap． 50k preset 100k inn．sp．
$100 \mathrm{k} / \mathrm{l}$ ．

## CABY MULTI－METERS

Inclusive of Test prods，Batteries，Inatra．Book． MODEL A－10．2I ohms／v． 10 rankee of volth， current，rewistance．E4．15．0，post $2 / 6$ ．
MODEL B－20．10s obms／v．on 0.57 ．and 2.58 4 ohnus／t．On（105．to 1000e．） 19 ranker． i6．10．0．peet． $2 / 8$ ．

## NEW METERS

0－500 Micro amp mec．yino ．．．．．．19j－ $0-1$ ma m．c． 2 in ．
0－30 ma m．c．2ho．
0－300 ma m．e．2，in．


## RECORDING TAPES

5 in ． $600 \mathrm{ft} ., 18 / \cdot ; 5 \mathrm{in} .900 \mathrm{ft}, 18 / 6 ; 58 \mathrm{in}$ 1200ft， $23 / 6$ ； 7in．1800ft．，35\％．

## ＂MYLAR DUPONT＂TAPES

6 in ．1200it．， $37 / 6$ ；7in．1800it．，44／－：7in．2400ft， 60／－．All post $1 /$ ．

## CAR RADIO AERIALS

Telencopio．Heavily chromed．couplete witb twountiag equiproent．22／6．

## CAR RADIO KITS

Tranulstorised，M／L waves．No नibrator． 12 volt stall．Can be built for 1212.10 .0

## BBC／ITA AERIALS

3 Element combined，31／6．Post 2／6．
LINE OUTPUT TRANSFORMERS
Most tyjues avallabic．State make of recelver． S．A．E．ai enguirtes．

## TRANSFORMER FILAMENT

230v．Primsry：6．3v．－1．3 \％．，6／8；210v．－240v． Prinary：6．3v：－2．75a．．8／\％Post 6d．

## J．B．DIAL ASSEMBLY

Type SLIB，General Purpoe slide Rule Drive for FM／VEF unis，8hort Wave Converters，etc． Printed in three colours on aluminium with a $0-100$ scale．Provision is made for Indicidual callibrations．Trave！of pointer $4 / \mathrm{in}$ ．，scale plate OUR PRICE $7 / 6$

## MIDGET MAINS

230v．Primary．220v．-30 ma ．and 6．3v．1ja． Standard， $9 /$－
As alove，Drop through mounting，semi－shrouded， 10／6；Fully sbrouded，11／．

## TOOLS

TWIST DRLLLS．Set of 9 ta plastic cabe， $1 / 10-\frac{10}{} \mathrm{in}$ ． $3 / 8$ Set．
SCREWDRIVERS．Reversible blade，Pbillips No． 1 and 2．Clear plastle bandle．2／8．
RADIO PLIERS．Gun linish．ouin．，3／11；Bjin．，4／6．

## MINIATURE

## DYNAMIC SPEAKERS

A must for all build－it－yourse if hams．As supplied with all ourrent wansistor kite．Can also be adapted for bomes phones of inter－cons．ail．diameter resistance 70 ohins．4／8．Pois bd．

## CRT HTR ISOLATION TRANSFORMERS

New improved types，low capacity，minall aize and ag terminatud．Prim．AC $200 / 250 \mathrm{v}$ ．Becondaries
 or 13 v ．Tubes． $12 / 8 \mathrm{each}$ ．

## 13 CHANNEL TVs

TABLE MODELS, FAMOUS MAKERS, Complete with all valves and tuber. uequalled in value.
AMAZINGLY POPULAR-IDEAL SECOND SETS




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 DIRECT FROM OUR FACTORY Due to the increasing demand for our wide range of CRTF REGUNNED TUBES have new Guns Gettorn. Aquadiac Coating Baseg eta.TOE QLALITY (iUAIAANIEEI)
Carriage and Insurance 12/6
CRM92. MW2\&-\%. MW23-14, MW22-14C, MW22-1. MW2L-18 GRM191 CRM121A 12KPA, 121K, C12B, CRM121, CRM121A CRM121B, CRM1ER, CRM128, MW31-14C.
MW31/16, MWB1/18, MW31/74 $1416,7201 \mathrm{~A}$ C14FM. CFM141, CRM142, ORK WRG/ MW43-64 MW45-64
F4U1A. AW4s-30. C14BM, C17RM, CRM151
CRM152A CRM182b. CRM15す, CRM173. MW4 S-80. MW41-1
A W53-80, CRM212 MW56-20, MW53-80 REVA

12 MONTHS RAND NEW

TYPES
M W31/74
C4-15
£4-15
M W36/24
£5-15
MW43.64
£6-15

| REVAZUUMED | REGUNiNED | BRAND NEW |
| :---: | :---: | :---: |
| \}. £1-10 | '£3 | TYPES |
| \} $£ 2$ | ¢4 | £4-15 |
|  |  | M W $36 / 24$ |
| ¢ $£ 2-10$ | £4-10 | ¢5-15 |
| $\left\{\begin{array}{l}\text { ¢ }\end{array}\right.$ | £5-5 | MW43'64 |
| ऽ £4 | £6-10 | £6-15 |
| es not listed | vailable. | .E. enquiries. |

STAAR GALAXY SPARES
Huge quantivy (Autochangers). S.A.E. enyurines

* Record Players GARRARD 4 -speed $4 / 5+6$ GARRARD RCl20I) Mk. II B.s.R. (けAlf) Antochanger EMI 4 -speed stereo

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##  <br> 4-SPD. RECORD PLAYERS

Lateen B.S.R. TU9 Turpranke, tokether with 3 Lightreight staar Galaxy dual sapphire crystal turnover mek-ip head. Truly amazing value $£ 3.10 .0 \quad$ Carr.
galaxy PICEUPs, as above. only 19/6.
External ITV Converters with power pack
Haramered finish. Very compact. Gain and


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RED SPOT Now only $3 / 6$. WHITE SPOT 4/6 DIODES

Carrage 41- + | $\because$ | $\because$ | ${ }^{85}$ |
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| 8 | 19 |  |
| 17 |  |  |

89
${ }_{8}^{85} 19$
19

| 57 |
| :---: |
| 6819 |
| 6 |

${ }^{\text {f } 6} 19{ }^{6} 19$
${ }^{66} 15$ dro. GEX44 t5ne, $3 / 8$.

## SPECIAL OFFERS <br>  UA8 Stereo Changers <br> B.s.R. <br> Unin ster with quality neren cartsulte. £6.190

## MAINS AMPLIFIERS


 Ideal for record plasere, P.A. wnrk, et

## Loutspeakers

TOP MAKES - MANUFACTURER FRESH



 HF1n12, 99/9: 12 2n. ' 'll, weet Field, 2786.


PM SPEAKERS Surplus 3 ohm | Tented. |
| :---: |
| $\substack{\text { gintin. } \\ \text { Rin. }}$ |
| $9^{\text {top }}$ |

For Chargers, ne lenıum, rull wave, 12 volt, 3 -4 anuph
100 RESISTORS 6/6 100 CONDENSERS ..... 10\%
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3 pF VALVES ALL GUARANTERD 3 MONTHS

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## NEW LOW PRICES <br> GUARANTEED 3 MONTHS

## FREE TRANSIT INSURANCE, All valves are new

 or of fully guarantect ex-linvernment or ex-equip ment ormin. Ftatisiact hon of Mones back Guarapthee on usmis if returnefl unused withm 14 daỹ


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There are no hatter value-for. money Tape Recorders on the market-if you can't call and hear them send S.A.E for fully descriptive leafiets.

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Each Model Incorporates the
hatiy succesful HF/TR3 Ampl-
fler (described opposite) thus
ensuring truly "H1-Ft" record and playback facilities.

BEFORE YOU BUY -YOU SHOULD HEAR THESE

## TAPE REGOBDERS

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COLLARO 'STUDIO"' TWINTRACK 3-speed Deck
H.P. Terms: Deposit $£ 2.18 .0$ and 12 months of $\mathbf{3 9 . 1 0 . 0}$ £2.1\%.11.
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TAPE AMPLIFIERS and PRE-AMPLIFIERS PRESENTED FROM MULLARD DESIGNS
MODEL HF/TR3 TAPE AMPLIFIER
Mullard Type "A" design)
A very hizh quality Amplitier Incorporating 3 -speed treble equallsation, using the latest INDUCTOR. FOR COLLARO TRUVOX-BRENELL-WEAR TE or MOTEK Tape Decks former. Includes Pormer Includes separate ${ }_{\text {PARTS }}^{\text {MAT }}$ E12.15.0
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## SPECIAL 'COMBINED ORDER' PRICES

(a) The COLLARO mull Type "STUDIO" TAPE DECK and our Unit Assembled and Tested
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As (c) but Type "C, and compliths eq.ili. The TRUVOX Mk. VI DECK and the assembled Type "C" Pre-amplifier and Power Unit.
(f) As aboveposit e8.0.0, and 12 months 82.18 .8
(g) The BRENELL Mk V DECK and the assembled Type "C" PRE-AMPLIFIER and POWER UNIT
(h) As (g) but TVpe "C'" as complete KIT OF PARTS
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(Carriage and Insurance on above quotes $10 \%$-extra.)
EACH OF THE ABOVE CAN BE SUPPLIEDIN A PORTABLE CASE

## THUS FORMING A COMPLETE PORTABLE PRE-AMPLIFIER FULL DETALLS ON REQUEST <br>  TRUVOX and GARRARD TAPE EQUIPNENT MICROPHONE-AN $1,200 \mathrm{ft}$. SPOOL E.M.I. TAPE-ALL FOR

ENABLES THESE OUTSTANDING PRICE REDUCTIONS
 The "MODEL HFIG2R' PORTABLE TAPE RECORDER (Original Price (33.0.0) FOR ONLY 22 gis. H.P. Dep. $£ 4.14 .0$. 12 months
£1.13.8. (Carr. \& ins. 101 -extra). €1.13.9. (Carr. ${ }^{\text {Incons. }}$ IO1-ertra). Incorporates THE LATEST
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is entirely based on the very successiul MULILARD TYPE very
to operate the GARRARD DECK. PRICE DN
CLUDES SUPPLY OF THE GARRARICE INMAGAZINE and a $41 n$. SPOOL of DOUBLE PLAY TAPE, COMprises a Twin Track Recorder operatinz at 3yin/sec, snoed and weighs only 22 lbs . Outstanding features are excellent nerformance and simplicity of operation.

TAPE PRE-AM1LIFLER-
The "H1-Fi" link to add full tape recording facilities to High Fidelity home installations. In-
CORE PUSH PUILL OSCILLATOR and 3-speed treble equalisation by FEROXCUBE POT CORE INDUCTOR FOR WEARITE-COLLARO-TRUVOX-BRENELL or MOTEK TAPE DECKS. ncludes separate power supply KIT OF £14.0.0 H.P. £3.8.0 Deposit and $\underset{\text { PARTS }}{\text { E17.0.0 }}$

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\text { (Excluding power unit £11.15.0 and } £ 14.10 .0 \text { respectively). }
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(a) COMPLETE KFT to bulld the HFFTR3 Amplifier. $\mathbf{£ 2 5 . 1 0 . 0}$ (b) As above but with HF/TR3 supplled A.SSEMBLED and TESTED

H.P. Deposit $£ 5 \cdot 16.0$, and 12 months £2.2.6
with the MK. IV COLLARO "TRANSCRLPTOR" TAPE DECK
£30.15.0
( $£ 1$ extra if we are required to wire up Deck Switch Banks.)
(d) As above but HF/TR3 suppLed ASSEMBLED and £34.10.0 1
(e1 extra if we are/required to wire up Deck Switch Banks.)
(e) COMPLETE KIT to build the HF/TR3, together $\mathbf{~} \mathbf{3 6 . 0 . 0}$
(f) As above but HF/TR3 supplied ASSEMBLED and $£ 39.10 .0$
H.P. Deposit 97.18 .0 and 12 months $£ 2.17 .11$.
£41.10.0
FIER with the BRENELL Mk. V TAPE DECK
) As above but HF/TR3 supplied ASSEMBLED and
£45.0.0
(d) THE WEARDTE \&A.O.O, and 12 months e3.8.0. tested HF/TR3 Ampliffer including WEARITE Head Lift Transformer
£55.0.0
Carriage and Insurance on each abnve in 10/- ext.o.
Atractive PORTABLE CASE is available to accommodate the TRUVOX or COLLARO TAPE DECKS and we offer it, together with ROLA/CELESTION 10 x 6 in . LOUDSPEAKER-ACOS CRYSTAL The "MODEL TK/MklV"
PORTABLE TAPE RECORDER (Original Price $\{49.10 .0$ )
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Incarporates tha Trivoz Mkiv
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In. Loudspeaker and the Truvox
Type "K". Amplifier specifically developed by Truvox Lid to correctly operate their MkIV Tape Deck. As a resule we are able to present a highly persected tape recorder provid-
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## Practical Wireless



YOL. XXXYI No. 649 MARCH, 196I


## ANOTHER FREE BLUEPRINT

IN the next issue of Practical Wireless, the first of a new serics will appear on constructing the "P.W. Roadfarer". This is an A.M./F.M. Recciver which is transistorised throughout. Its advance design allows it to be operated either from its own internal batteries or direct from the domestic mains supply. Printed circuits have been used and the construction of the receiver should therefore possess lew difficulties even for the inexperienced constructor. The set is constructed in four main sections: the A.M. tuner; the F.M. tuner; the audio amplifier and the main power supply unit. The comptete receiver is housed in an attractive plastic cabinet of small size which contains a ferrite rod aerial for A.M. reception and also a telescopic aerial for F.M.

Each cony of the April issue will contain a free blueprint of the "Roadfarer"-normal price 5 s . Od.-which will give the complete circuit diagram and components list and other information necessary to build the set. Be sure to order your copy NOW.

## The P.W. Signal Generator

THIS issue contains the first article of a new series on the progressive construction of a wide range Signal Generator which will provide all the facilities required by the amateur radio enthusiast and will stand comparison with many of the designs for the amateur which are now on the market. The series is written by E. V. King-the author of two nopular, constructional series which have previously appeared in this magazine. Whilst the performance of the Signal Generator is excellent, great care has been taken in the design to minimise the cost of construction and together with the progressive construc= tional system used, enables the cost of the unit to be spread over as long a time as the consiructor desires. The Generator is constructed in four classis and if required one or two chassis may be omitted if the constructor has no use for the facilities which they would provide.

## Talking Books for the Blind

TALKING books for the blind have been used for some years but now a new library of tape recordings is being accumulated to replace the long-playing gramophone record system. A revolutionary type of play-back machine will be used which possesses a number of basic advantages over the present dise reproducer, the result of almost twenty-five years' research and development by the R.N.I.B. and St. Dunstan's. Amplifier, loudspeaker and drive mechanism are housed in one compact case, controls have been reduced to one On'OffVolume switch and one Siart/Stop lever.

The truly unique feature of the machine, however, is the tape cassette. Litlle larger in size than an average letterpress novel, this can contain up to twenty hours recorded speech. This has been achieved by the use of half-inch tape and recording on it not two tracks, as on domestic machines, but no fewer than eighteen tracks, which reduces the number of times a tape must be changed and so simplifies matters for the user.
 Our next issue, dated April, will be published on March 7th

# Honnd the World of Wircless 

## POTENTIAL AND CURRENT NEWS

## Broadcast Receiving Licences

THE following statement shows the approximate number of Broadcast Receiving Licences in force at end of November, 1960 , in respect of wireless receiving stations situated within the various Postal Regions of England, Wales. Scotland and Northern Ireland. The numbers include Licences issued to blind persons without payment.


## Factory Extensions

Factory extensions at Luaton Road, Dunstable, scheduled for occupation in March. 1961. are announced by Erg Industrial Corporation Lid. The increased floor space will make it possible to increase the output of the wire wound Resistor division. Additionally, new and modern equipment for the now established Erg Transformer division will occupy a large proportion of the new extension. with an enlarged airconditioned, constant-temperature laboratory. Test facilities covering full service requirements will be in oncration in ensure the highest degree of reliahility.
Erg Industrial, already serving the Electronic, Industry with "Custom-built" special products. intimate that strong emphasis is being given to all immediate expansion of this service. for which highly qualified design engineers have heen engaged and are solely devoted to this "Custom-built" division.

## Dopplers and Automatic Direction Finders

$\mathrm{A}^{\mathrm{F}}$FTER an exhaustive evaluafion, the Ministry of Aviation has selected the Marconi Doppler Navigation Type AD 2300 B and the Automatic Direction Finder Type AD712 for use in their new


Transistorised plug-in logic units form the basis of the electronic signal interlocking developed by Mullard Equipment Lid. for British Railways (see page 967).

Argosy C. Mark 1 aircraft for R.A.F. Transport Command.

The doppler order is worth $£ 552.000$ and the automatic direction finder order $£ 100.000$. Ancillary equipment and spares are included in both contracts.
The AD23008 comprises the hasic AD2300 sensor. the 5452 compucr and the 5453 display unit. The computer processes ground speed and drift angle from the doppler sensor: ogether with heading information from the aircraft compass system. and pecsents to the pilot the aircraft's displacement from selected track and distance to go to the turning point. It also feeds steering information into the flight system and/or autopilot to enable the selected track to be maintained.
The AD712 autonatic direction finder is a crystal-controlled ADF embodying a true and positive decade luning sysiem and a fixed loop aerial sysiem. Frequency selection is by switch: no luning whatever is required. This equipment is in wide use by many airlines.

## Battery Firm Changes

'TO achieve closer liaison hetween their technical, production and sales departments.

Chloride Batteries Ltd. transferred the major portion of their sales headquarters from London to the head office and the principal manufacturing centre of the company at Exide Works, Clifton Junction, near Manchester, on January 2. 1961.

Mr A. C. Stewart, the present sales manager, who is approaching retirenient age, will not be moving north. He will be succeeded by Mr. M. A. GriffithJones on completion of the move. hut his services will be retained by the company in the capacity of consultant until hic retires.

## Export TV for Design Index

THE latest Ekco export television models - the 17 in . models T737 and T738-have been accepted by the Council of Industrial Design for "Design Index". The "Design Index" already includes nearly sixiy Ekco products and is Britain's record of well-designed goods from which exhibits are chosen for display in the Design Centre.

## Tope Recorders for BBC

T'HE RRC is purchasing two three-track portable and one twin-track console Philips mag-
netic tape recorders for use in engineering research work.

Negotiations for the order were carried out for Philips by Mr. Terence Perkins, the company's tape recorder section manager.

## Silicon npn Diffused Junction Power Transistors

${ }^{5}$ THE Semicondactor depariment of A.E.I. Radio and Elcerronic Components division entered the silicon power tramsistor tied with the introdaction of three l.diswan Mazda npu dillused jumcion type devices. Types XC 703 , XC'713 and XC723 are imbended for a wide vartiely wimbustrial applications in cquipment opera* ting at comperatures ranging from $-65^{\circ} \mathrm{C} 10+175^{\circ} \mathrm{C}$. Ihay are particularly uselul in powerswitching cirenits, oscillator regulator and pulse amplilier circouts, and as Class A and Class B pushpoll audio and servo amplifiers:

They are charaterincel by an extremely low saturation resistance, high current and power dissipation ratings, high Beta, at high current, and excellent high temperature performathe up ${ }^{\text {(1) }}$ $+175^{\circ} \mathrm{C}^{\circ}$ A! $100^{\circ}$ ( $^{\circ}$ they wller a range of dissipation of lrona 210 30W. They are available in slandard JEI)IEC: carts.

## Direct View 3-D Picture Process

1FUlIL-(COLOUR 3-1) pictures that can be "blown up" to $40 \mathrm{in}, \times 30 \mathrm{in}$, are now being produced for viewing without coloured spectacles or a stereoscope by a new process sponsored by Deep Vision LId.

This process is based on the familiar grid system in which each of the viewers eyes soes the pieture through at different set of "windows" - hamdreds of extremely narrow vertical slits between invisibly dine upayne lines.

But apart from their size, the Deep Vision picaures can also be brought forward from their frames or set back in perspective —or both. And as in the calse of the equally familiar lenticular system (in which verlical corrugations break up the picture into two eye-tields), they ean also be made to "change".

Two more advantages are that any number of identical prints can be made in fast colours. and that as 2-D colour pictures can be made simultaneously, magazine advertising. for example, can be precisely matched in 3-1) for illuminated display.

The process involves the use of - specially-designed camera which, although bulky and heavy, has no single awkward component, and so can be dismantled and reassembled in a matter of minutes for location wark.
A shot takes about five seconds to complete, but considerable subjective movement can be tolerated withont the pieture being spriled.

## Talk-back System for New Tamar Bridge

Tilicil:RS working high up, on and between the 250f: towers al the new Tamar Bridge - Now being huile by (leveland Bridge and tinginsering ( 0. . 6 provide : new road link between Devon and (ornuall-beep in touch with the main control station at one end of the bridge through çupment supplied by Radio Componcols and Special Prodack Depatment A.Fi.t. Kation and Flactronic ciont prosents livision.

- ${ }^{2}$ lice twin towers of the bridge are l.looft apart and the distance herween the top of each and twa grommal stations is 500 ft So that commmbication was reguited over a total distance of about 2.000): To provade this
a transistorised loudspeaking intercommunication system has been supplied. A ten-line master unit, of a type specially developed by A.E.1. to work over a post office line and therefore eminently sultable for this application, is installed in the constrtictor's main utfice at olte end of the bridge.


## British Railways to use Electronic Signal Interlocking

As patt of the railways nodernis:aion programme, the British Tramaport (ommission has awaraled at contract to Mallatid Eynipment I.td. for the development of an electronic signal intertoching plant. The eyuipnemt will be the first plant thing electronic interloching to be ascal on Britialn Railways and will replate the mechanical interlocking on the lever frame at Henley-on-Thames Western Region signal bor. It will be intalled during the early summer


Recenily the company was able to denwnstrate 10 British Railway signitl engiteers a working muntal of a donitate junction, completely controlled by the new electranic syatem. The technique has the substantial advantage that it employs solid state eleo tronic eircuitry throughout.


Final testing in progress at the New Plymouth terminal station of the New Zealand microw'a'e telephone network (installed by STC). The four equipment cubicles on the left are the transmitter and receiver equipments for the "working" and " standby" radio chummels. The supervisory VHF radio equipment is in the background.

## A MIDGET



The componem layout of this amplifier.

By V. E. Holley

## 9

HIS amplifier was built originally for a portable record player. Though modest in size, weight and cost, it has an output of four watts and a very acceptable standard of reproduction.

## The Circuit

Fig. 1 shows the circuit; it is largely conventional. The valve, V1, is a resistance coupled pentode voltage amplifier having in its grid circuit

## THIS 4W AMPLIFIER IS ESPECIALLY SUITABLE FOR PORTABLE RECORD PLAYERS

the volume control, VR1, and a simple top-cut tone control, Cl and VR2. Matters are so arranged that in the minimum cut position, the tone control has practically no effiect on the response, but there is sufficient range to enable the surface noise from the older $78 \mathrm{rev} / \mathrm{min}$ shellac records to be reduced to inoffensive proportions. R3 is the bias resistor which is by-passed in the normal way by C3; R4, inserted between the


Fig. 1a.-The circuit diagram.
lower end of these components and chassis, provides a point for the injection of negative voltage feedback derived from the output transformer secondary.

## Inverter

The amplified signal from $V_{1}$ is passed to $V_{2}$, a triode, 6 C 4 , arranged as a phase inverter with

Fig. 1b (right).-The valve-base connections.


Fig. 2.-Constructional details for the chassis.
equal anode and cathode loads of 47 k . Two signals, 180 deg out of phase, are thereby produced for the push-pull output stage. A cathode bias resistor of 2.7 k and a grid leak of 1 M . returned to the junction of the bias and cathode load resistors, complete the stage. Heavy negative current feedback ensures excellent linearity though it limits the gain of the stage to 0.9 on each side, or 1.8 times overall. No de-coupling is necessary between V1 and V2.

## Output Stage

Two miniature output pentodes. EL91, are employed in this stage with 250 V on anodes and screens. They should be reasonably well matched but exact matching is not necessary as the common bias resistor tends to equalise them. The optimum load, anode-to-anode, is 24 K , so that for adequate low frequency response, an output transformer with unusually high primary inductance is required. Such transformers are not readily available but the difficulty can be overcome by using one of lower inductance and making good the low frequency attenuation by negative feedback. Several small push-pull transformers are available commercially with a primary inductance of 40 H
or so and affording ratios of around 80 or 90 to 1 for operating a $3 \Omega$ speaker. As the D.C. resistance of the speaker is rather high, there is some loss of volts. at the valve anodes and the resistor R13 is included in the supply to the screens so that the voltage here will not rise above that at the anodes.

## Negative Feedback

Heavy negative feedback is taken from the output transformer secondary via R14 and C8, to the cathode circuit of V1. The inclusion of C8 reduces the feedback at the lower frequencies and so allows the gain of the amplifier to rise in the region in which attenuation is introduced by the output transformer. The values of C8 and R14 will need adjustment according to the characteristics of the output transformer. It is suggested that the amplifier be completed with R14 equal to $330 \Omega$ and $C 8$ omitted, and that the best combination should then be selected after a listening test under service conditions. The optimum value for R14 will lie between 200 and $600 \Omega$ and for C8 between 2 and $8 \mu \mathrm{~F}$. The circuit is very well behaved and none of these combinations will cause instability.


The completc amplificr.

## Power Supply

The mains transformer should be of the "miniature" type and should have a half wave secondary winding capable of supplying 35 mA and a $6.3 \mathrm{~V}, 1 \mathrm{~A}$ heater winding. A contact cooled rectifier is adequate, and mains smoothing is provided by C9 and C.10 in conjunction with R15. The smoothing is not lavish, but push-pull plus negative feedback reduces the hum to a very low level. Some additional smoothing is necessary for the supply to V1 and V2 and is provided by 129 and C7.

## Construction

The prototype was built on a chassis of 18 s.w.g. aluminium sheet measuring $7 \frac{1}{2} \mathrm{in}$. $x \quad 2 \frac{3}{3} \mathrm{in}$. $x 2 \mathrm{in}$., a plan of which is given in Fig. 2. There is nothing critical about the lay-out except that to prevent inter-action between them, the mains and output transformers must be mounted so that their magnetic axes are at right angles. There is a little difficulty here if both are of clamp construction but it can be overcome by mounting the mains transformer on its side in the "dropthrough "position and securing it with a strap of 18s.w.g. aluminium across the top, the normal mounting lugs being bent inwards out of the way. This method was adopted in the prototype as will be seen from the illustration. A detailed wiring diagram for the amplifier is given in Fig. 3.

## Components

No special components are necessary but it is an advantage if the resistors and capacitors are on the small side. The resistors can be $\frac{1}{6} \mathrm{~W}$ except R9 and R12 (克W) and R15 (1W). R6 and R8 must be a matched pair as also must be R10 and R11. If matching facilities are not available, close tolerance components must be used ( 5 per cent or better). All the capacitors should be 350 VW D.C. except C3 and C8 for which 25 VW will be adequate.

## Testing

When construction is complete and before switching on, a check should be made with a meter between Cl0 and the chassis to see that there are no shorts in the H.T. wiring. If all is well the power can be connected and measurements taken of the voltages on the output valve anodes and screens. It should be between 240 and 250 V , measured between these electrodes and the cathodes, and some adjustment may be necessary to the value of R15 to make it so. depending on the output from the mains transformer. Likewise, it may he necessary to adjust the valuc of R 13 according to the voltage drop in the output transformer primary. If there is instability, the connections to either the primary or secontary of the output transformer should be reversed so that the feedback becomes negative. All that now remains is to decide by trial and error on the most suitable values for R14 and C8. It should perhaps be mentioned that if the amplifier is operated without feedback, the reproduction will be very

## COMPONENTS LIST

Resistors ${ }^{1}{ }_{4} \mathrm{~W}$ unless otherwise stated):
R1-1.2M
R2-0.27M
R3-2.7k
R4-27
R5-1M
R6-47k
R7-2.7k
R8-47k
R9-15k (1 $\mathbf{W}$ )
R10-0.47M
R11-0.47M
R12-600 ( $\frac{1}{2} \mathrm{~W}$ )
R13-1k
R14-330
R15-470 (1W)

Potentiometers:
VR1—1M log.
VR2-100k log.
Valves:
V1-EF91, B7G base and screen
$V^{2}-6 C 4, B 7 G$ base.
V3 4$\}$ EL91, B7G hase
Mains transformer:
Miniature, 250 V half wave, $35-40 \mathrm{~mA}, 6.3 \mathrm{~V}$, 1A
Output transformer:
Push-pull, high inductance primary (see text). Ratio 90 to 1 for $3 \Omega$ speaker or 40 to 1 for $15 \Omega$


Fig. 3,-The component wiring details.
high pitched and, because the de-coupling is not generous, there may be a slight tendency to
 instability. These faults will disappear when the feedback circuit is connected.

Fig. 4.-The pick-up compensation circuit.

## Operation

If the amplifier is to be used with a pick-up, it will be necessary to provide some compensation for recording loss. While this could easily be incorporated in the chassis, it will usually be more conventent to fit it at the pick-up so that it is eliminated when the amplifier is used for other purposes. The output of the popular type of crystal pick-up rises in the low frequency region and it will generally be satisfactory to add external compensation at the rate of 3 dB per octave below aboui $250 \mathrm{c} / \mathrm{s}$. This may be done with the circuit of Fig. 4.

## Heartbeat Timer

An electronic heartbeat timer with no external wiring can be sewn into a patient's chest or near his heart because it has self-contained batteries. The circuitry is in a plastic envelope about $2 \frac{1}{2} \mathrm{in}$. in diameter, sin. thick.
The timer supplies a stimulating pulse at regular intervals to keep the heart beating regularly, where otherwise it might beat dangerously out of time, slow down or stop entirely.

This heart timer will run for five years, and has been tested on dogs for periods up to 100 days. Dr. Wm. Chardack, of the Veteran's Administration, and electronic consultant Wilson Greatbatch. of Buffalo, N.Y., U.S.A., developed the device.
Meanwhile, it was resported that RCA in America developed a heart timer using a generator strapped around the patient's chest which induces power into a device sewn inside the chest cavity.

## High Power Klystron Installations

One of the 50 cm commercial radar installations using the English Electric Valve Company's K347, 600 kW Klystron is well on the way to completion at Kastrup. Copenhagen's new international airport, which is expected to be operational some time this month. The K 347 is the first high power klystron to be designed specifically for use in commercial radar and in this installation two klystrons are used with a Marconi S264A radar comprising two transmitter/receivers feeding an aerial system on a main/standby basis.

The K347 is an external cavity tuned klystron operating in the 585 to $610 \mathrm{Mc} / \mathrm{s}$ frequency range and repeated tuning operations can be made, to meet changing operational requirements, without stressing or damaging the tube.
 C.W. TRANSMITTERS

By G3OGR

(Continued from page 877 of the February issue)

$7 /$Using a doubler, make sure that the crystals chosen are of such a frequency that operation is always in the permitted bands. Provided this is remembered, a doubler will make available other frequencies, with a limited number of crystals. It is also in order to use a further doubler. This is often done to permit three band working from one crystal, for example.


Fig. 6.-Oscillator and multiplier stages.

## Power Supplies

The circuits shown have great latitude in power supply requirements. Almost any power pack available may be used. This item can thus be arranged to suit the output wanted. Even a small receiver type supply will allow 10W or more input to a single valve transmitter.

A typical separate power pack is shown in Fig. 7. The mains transformer, choke. and rectifier should be chosen to suit the output voltage and current desired. For inputs of up to 50 W or so, a $400 / 0 / 400 \mathrm{~V}$ H.T. secondary rated at 150 mA , with a 5 R4GY rectifier, will be satisfactory. The choke current rating should at least equal the current drawn. The working voltage of the smoothing condensers should be rather higher than the actual H.T. voltage.

A fuse is an added safeguard against short circuits or similar defects. A 100 mA or 150 mA fuse will do for small power packs, with 250 mA for latger supplies.


Fig. 7.-Power supply circuit for the transmitters.

If the power pack can be disconnected, or its circuit is interrupted by keying. a 100 k 5 W bleeder resistor should be connected across the H.T. supply, to discharge the condensers. In some circuits the screen grid resistors will act as a bleeder.

It is usually necessary to insert a switch. so that the valve heaters may be run without H.T. being applied. When initially testing a circuit, it is advisable to apply reduced H.T. voltage, until resonance has been found. A dropping resistor may be used for this purpose.

## Key Clicks

Crystal oscillators usually key well. but if key clicks are produced, the cathode or key by-pass condenser may be changed in capacity. Using a rather large capacity will tend to cause "tails" to the characters, but this depends on the current drawn by the valve.

If a condenser alone is insufficient, a choke may be added in series with the key. Another condenser may then be joined directly across the key contacts.
Many operators have favourite methods of keying. The key may interrupt the cathode circuit, or may be in the H.T. negative circuit. An alternative is to use a separate transformer to obtain H.T. only, and key this in the primary. The rise and decay time can then be adjusted by modifying the smoothing condenser capacity. When-keying mains and other high voltage circuits a relay is preferable, for maximum safety.

# A BANI II Pre=Amp 

A USEFUL ADDITION TO RECEIVERS FAR FROM THE TRANSMITTER
(Continued from page 874 of the February issue)

WHEN construction of the preamplifier is finished, all wiring should be checked carefully. The unit may now be switched on and if possible the H.T. and L.T. should be measured without the valve in position. If all is in order the unit should be switched off, the valve inserted and the unit switched on again. As a check, the H.T. and L.T. voltages may again be measured. Plug the downlead into the input socket and connect the output to the receiver aerial socket.

The dust cores in L2 and L4 should be altered in position for best results and finally the positions of the two coupling coils L1 and L3; movement of these coils may necessitate altering the position of the slugs again. If the slugs are loose in the coil formers they should be removed and a piece of thin elastic inserted down the threads of the former and the slug screwed in again. When the best positions for the slugs and coupling coils have been found. the coupling coils may be lightly cemented in position. (A tool for adjusting the position of the cores can be made from a plastic knitting needle.)

## Alternative Coupling

If the incoming signal is very weak, it may be


Fig. 4.-The above-chassis layout of the unit.

found that the signal from the pre-amp is less than that from the aerial direct. Whilst this may be due to L2 and L4 resonating at the wrong frequency, it is more likely that the coupling between the aerial and L2 and the receiver and L. 4 is not optimum. Various alternative methods of coupling may be tried instead of using the coil L1, the wire from the aerial input socket may be tapped directly on to L2 and the optimum position be determined by trial. The effect of including a small capacitor-2-10pF-may also be tried.

Matters are easier as far as L4 is concerned. The effect of omitting L3 and employing a small capacitor between the anode of V16 and the inner conductor of the output coaxial cable is certainly worth trying. The value of this capacitor may be again $2-10 \mathrm{pF}$. If good results are obtained with the initial arrangement the number of turns on L1 and L3 may be altered in an endeavour to secure better results.

## Final Adjustments

The condenser Cl , which is part of the neutralising network is set for optimum results. The turns of the inductance L5 may be opened out or closed up to secure best results. Accurate alignment of the circuit is best carried out by using a sensitive meter to monitor the voltage of the limiter grid of the F.M.
receiver. A resistor of about 10 to 47 k should be included at the receiver end of the meter lead. (Note that the positive lead of the meter must be connected to the chassis of the receiver and the negative lead-through the stopper resistance just mentioned-to the grid of the limiter valve.)

If it is desired to avoid the losses which may be
associated with change of cable at the input socket this socket may be omitted and the aerial downlead wired directly to L1a.

## Instablity

Owing to the high gain of the pre-amplifier it is possible and, if construction is not particularly well carried out, likely, that instability in the pre-amplifier will be experienced. In curing this instability the positions of the output and input leads relative to each other are critical and re-arrangement will often effect a cure. The decoupling capacitors can likewise have a great effect upon the performance-particularly C6 and C7. In the event of instability, it is worth while to experiment with these condensers by putting others in parallel, perhaps connected to different earth points, until the instability is cured.

## COMPONENTS LIST

Registors: Capacitors:
R1-82k , CI-2 to 10 pF , bee-R2-82! . hive type (or similar).
R3-100k 10per cent. C2-2pF.
R4-180』. C3, C4, C5, C6, C7, R5-100k 10 per cent. 1000 pF , min. ceramic. Valve: PCC89. C8-3pF ceramic.

## Power Unit Components:

Resistors, R6, 1 k or to suit H.T. voltage. Capacitors, C9 and C10, 8 to $32 \mu \mathrm{~F}, 350 \mathrm{VW}$. Rectifier, miniature contact-cooled type for 200 V r.m.s. input and rated at about 30 mA . Transformer-type used for converters, with tapped primary and 7 V secondary.
Chassis: see diagram (Fig. 4.)
Solder, wire, wooden case, etc.

## COIL WINDING DETAILS

| No. of turns |  |
| :--- | ---: |
| L1 (over L2) | $\mathbf{1}$ |
| L2 | $\mathbf{7}$ |
| L3/4 (over L7) | $\mathbf{2}$ |
| L4 | 6 |

Coils are wound on łin. small polystyrene coil formers with purple coded dust cores. No spacing is used between turns-the coils are closewound.
LS consists of about 15 turns of $\mathbf{2 2 - 2 6 s}$.w.g. enamelled or DCC copper wire wound to $\frac{1}{2}$ in. inside diameter.


| COMPONENTS LIST |  |
| :---: | :---: |
| Resistors: Ca |  |
| R1-82k. | CI-2 to 10pF, bee- |
| R2-82!. | hive type (or similar). |
| R3-100k 10per cent. C2-2pF. |  |
| R5-100k 10per cent. 1000 FH , CS, C6, C7, |  |
|  |  |
| Valve: PCC89. C8-3pF ceramic. |  |
| Power Unit Components: |  |
| Resistors, R6, 1 k or to suit H.T. voltage. |  |
| Capacitors, C9 and Clo, 8 to $32 \mu \mathrm{~F}, 350 \mathrm{VW}$. |  |
|  |  |
| 200 V r.m.s. input and rated at about 30 ma . |  |
| Transformer-type used for converters, with tapped primary and 7 V secondary. |  |
|  |  |
| Chassis: see diagram (Fig. 4.) |  |
| Solder, wire, wooden case, etc. |  |



## Housing

The finished unit may be housed in a wooden box which can be fitted or placed on top of the radio receiver.
Although a great deal of heat is not generated in the unit. adequate ventilation should be assured to avoid frequency drift. During the testing, the unit can be operated direct from the mains supply bit finally when it is fitted permanently to the receiver the mains supply should be taken direct from the set (after its mains switch) and then the pre-amplifier cannot be left switched on inadvertently.

## Bandwidth

The damping of the input tuned circuit is such that the coil resonates over the whole of Band II and no tuning is necessary. However, it may be found that the tuning of the anode coil is sufficiently sharp to render impossible a flat response over the complete band. In this event, a damping resistor may be connected across the coil L4 in parallel with the condenser C8. It may have a value of, say. 1 k to 4.7 k , depending on the damping required to obtain best results. This resistor may result in some loss of signal, and if reception is such that no signal can be "wasted." then it may prove necessary to employ variable tuning in either the input or the output circuit.

## Conclusion

Ift is worth while emphasising again that where signal strength is low a pre-amplifier such as this must be "tailored" to suit the conditions of the location concerned. To obtain maximum transfer of energy from aerial to first grid and from second anode to receiver will necessitate, in most instances, some experiment, but it will then be certain that the best use is being made of the available signal.

# Valve Bias Explained 

BASIC THEORY FOR THE BEGINNER

ACIRCUIT for a triode valve is shown in Fig. 1 and is representative of many audio-frequency amplifiers. At this stage let us consider the circuit purely from the D.C. point of view in relation to the resistive elements connected to the various electrodes.
The valve heater is energised either from a low voltage winding on the mains transformer or, as we will recall from the last two articles, direct from the mains through a voltage dropping resistor in series with the heaters of the remaining valves. With battery valves, of course, the heaters are powered from the L.T. battery.

It will not be out of place here to discuss in broad outline the action of a triode valve. The heater of the valve (see Fig. 2) is placed inside a hollow tube (the cathode) which is coated with a white powder such as thorium oxide. When the heater is switched on, it heats the cathode causing it to glow. The white powder as a result of the heating emits "electrons" which are tiny particles of electricity. These particles are charged negatively. Now negative charges are repelled by negative


Fig. 1 (left).-A simple basic circuit for a triode valve.
Fig. 2 (right).-The conventional circuit symbol for a triode valve.
charges and attracted by positive charges. Therefore, if the anode (made up of a metai plate) is charged positively with respect to the cathode, then the electrons emitted from the cathode are attracted to the anode and travel to it. This flow of electrons constitutes a current of electricity through the valve.

The grid of the valve consists of a mesh of fine wire through which the electrons from the cathode have to pass on their way to the anode. If this grid is charged negatively with respect to the cathode, then some of the electrons travelling to the anode will be repelled and the current through the valve will be decreased. It can be seen that the current through the valve depends on the potential of the grid with respect to the cathodethe current, increases as the grid becomes more positive (or less negative) relative to the cathode and decreases as it becomes less positive (or more negative).


Fig. 3.-A valve bias system using both a cathode resistor and a grid potential divider.

The name "valve" is derived from the way in which the grid is used to control the flow of current, just as mechanical valves are used to control the flow of water in pipelines.
If, in Fig. 1 we assume that the grid potential is constant the conditions in the circuit are steady and the valve can be represented by a resistor of, say, Rv ohms (as indicated in Fig. 1). Thus we may consider, that there are five resistors in circuit, Rd, Ra, Rv, Rk and Rg. Thus current flows through the valve of a value determined by ( $\mathrm{Rd}+\mathrm{Ra}+\mathrm{Rv}+\mathrm{Rk}$ ), neglecting Rg .

## Resistor Calculation

In Fig. 1 it is supposed that the H.T. line is 300 V positive with respect to the chassis (H.T. negative). The circuit constants may be arranged so that with 200 V at the junction of Rd and Ra , with respect to chassis, the current through the valve and resistive elements is $10 \mathrm{~mA}(0.01 \mathrm{~A})$.

In this case, the value of Rd can easily be computed by using Ohm's law, this is, $R d=(300-200)$ ).
$(0.01)$. This follows because Rd is called upon to drop 100 V , this being the difference between the H.T. line voltage and the required voltage at the junction of Rd and Ra. This works out to 10,000 ohms (10k). Thus, a voltmeter connected to the junction as shown in Fig. 1 would read 200 V . provided Rd is 10 k and the meter does not load the circuit in any way. In practice, of course, the meter acts as a load across the resistive elements and the reading as indicated is below that computed as governed by the extent of the loading.


Fig. 4.-A circuit using a common bias resistor to obtain bias for all valves-the bias lines are connected to the grids of the valves through decoupling networks, and the cathodes are connected direct to chassis.

The 10 mA of current is also flowing in Ra and Rk. The value given to Ra is usually decided by the operational factors of the circuit, as distinct from the D.C. or static aspect, and the same applies in some degree to Rk in the cathode circuit.
The value of Rg is usually very high, often in the region of 1 M . but as current does not normally flow in this resistor, there is no potential difference across it and the grid is at chassis potential. (Use Ohm's Law: $V=\operatorname{Rg} \times 1, I=0$, and therefore $\mathrm{V}=\mathrm{O}$.) This fact is essential to the understanding of the following explanation and one which often causes confusion or bewilderment in the mind of the beginner but is easy to grasp when the principle is understood. There is no current through Rg ; therefore there is no P.D. across it.
Let us suppose that Rk is 1,000 ohms, and that the circuit is set up as previously mentioned with 10 mA flowing. The volts drop across Rk is then equal to $(0.01 \times 1,000)$ which works out to 10 V . Or another way of looking at it is to multiply the current in mA by the resistance in $\mathrm{k} \Omega$, which in this case is $(1 \times 10), 10$ volts.
The polarity of the voltage across Rk is such that the cathode is 10 V positive with respect to the chassis or H.T. negative line, as shown on the circuit. The grid of the valve is connected to the H.T. negative line or chassis through Rg. This means that the cathode grid is 10 V positive with respect to the chassis and grid or stating it another way, the grid is 10 V negative with respect to the
cathode, always provided 10 mA of current is flowing in the 1 k cathode resistor.

The best way of measuring the bias voltage is by connecting a voltmeter across Rk. There will be the least error in the reading by adopting this method because the resistance of the voltmeter on the 10 V range is almost certain to be far greater than the value of Rk . If it is attempted to measure the grid bias by connecting a voltmeter between the grid and cathode, then current will flow through Rg by way of the meter and the reading will be considerably less than the actual voltage.

If ever we require to discover the value of Rk for any grid bias voltage then all we do is divide the value of grid bias required by the cathode current of the valve. This, of course, is simply $R=E / I$, and where $R$ is in ohms, I must be in amperes, but where $R$ is in $k \Omega$, then I must be in mA.

Where a pentode, tetrode of any other multielectrode valve is in use, then the current through Rk is equal to the total H.T. current of the valve (i.e. anode current + screen grid current, etc).

Other curious methods are sometimes adopted to secure a bias voltage. Fig. 3, for example, shows such a method. Here is an R.F. amplifier which uses cathode bias, having Rk in the cathode circuit, but in addition has a positive potential on the control grid. Rk provides some fixed bias, but since the control grid is also connected to the junction of the potential divided R1, R2 some of the bias given by Rk is counteracted.

However. this arrangement provides a degree of compensation, and the circuit is stabilised to some degree in the event of a reduction in emission of the valve-when the white powder on the cathode begins to emit fewer electrons than before, resulting in an increase of the apparent resistance of the valve.

## A Common Bias Resistor

${ }_{1}^{1}$ Some receivers use a common biás resistor and have all the valve cathodes returned to chassis. The idea is shown in Fig. 4. Here Rb is the common bias resistor and is connected between the centre tap on the secondary of the mains transformer and chassis. It will be realised that the total current of the receiver flows through Rb, and a P.D., depending upon its value, is developed across it.

If the total current of the receiver is 100 mA $(0.1 \mathrm{~A})$ and Rb is $100 \Omega$, then 10 V is developed across it (i.e., $E=0.1 \times 100$ ). The voltage across Rb is negative at the secondary tap with respect to chassis, so it can be used ideally as a negative bias. In some cases the potential is applied direct to the grid of a valve after passing through a suitable filter circuit. while more often it is applied across a potential divider, such as that made up of R1, R2 in Fig. 4. In this way two bias lines can be drawn off and suitably filtered individually.

Since the bias is usually applied to a high resistance circuit relating to the valve, R1 and R2 can be of fairly high value, considerably higher, in "fact. than the value of Rb . This means that they have little shunting effect on Rb , which can be computed as given above.
The full bias voltage ( 10 V in the case cited) is present on bias line 1. and the bias voltage on line 2 is governed solely by the ratio of R1, R2, as fully detailed in a previous article in this series.

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# QUARTZ CRYSTAL ETCHING 

RAISING THE FREQUENCY OF A QUARTZ CRYSTAL

By J. B. Dance

## J

 HE resonant frequency of a quartz crystal can be increased somewhat by reducing the thickness of the crystal. It is not easy to alter the frequency of crystals which have gold or silver foil electrodes deposited on them (especially if the crystal is vacuum sealed), as these electrodes are usually "spluttered" on to the quartz using special apparatus. It would be difficult for the amateur to redeposit the electrodes on to the etched crystal or to connect wires to the foils. Such crystals are used mainly for frequencies below about $500 \mathrm{kc} / \mathrm{s}$. Above this frequency, rectangular quartz plates are common; the electrodes may consist of two metal plates held in position by a spring. The frequency of such crystals can be increased fairly easily by the method to be described.
## Technique

There are two main methods by which the frequency of a quartz crystal can be raised, namely by grinding (lapping) or by etching. It is preferable to maintain the surface of the crystal flat to within $1 / 100,000$ of an inch and therefore grinding creates some problems. If the crystal is not maintained perfectly flat, its activity diminishes; that is, when it is used in an oscillator circuit, the amplitude of oscillation becomes smaller and the crystal may even refuse to oscillate at all.


Fig. 1.-An oscillator circuit, using a crystal.
After a crystal has been ground, there is a certain optimum amount of etching which will lead to maximum activity. Ideally the amount of etching is such that the frequency is increased by etching by about the same number of $\mathrm{kc} / \mathrm{s}$ as the square of the crystal frequency in $\mathrm{Mc} / \mathrm{s}$. Whilst this amount is not very critical, it is to be expected that a good commercially manufactured crystal will lose some of its activity during etching. This can usually be tolerated, however, in order
to obtain a crystal of the desired frequency; other* wise preliminary grinding will be required.

## Etching Materials

Quartz is chemically very similar to glass and therefore very few liquids will etch it away. Hydrofluoric acid has been used for etching, but the average person is most strongly advised to avoid the use of such an extremely dangerous liquid. A solution of ammonium bifluoride, whilst much safer, is nevertheless quite dangerous. It can be ordered from most chemists at about 7 s . 6 d . for $\frac{1}{2} \mathrm{lb}$. A solution containing about 1 or 2 parts of the solid in 5 parts of water should be made in a plastic container. It is quite satisfactory to place the liquid inside a sheet of polythene which is put in the mouth of a wide-necked bottle or on a "tin-lid." It is wise to use the ammonium bifluoride solution out of doors and to keep a bucket of water nearby in case the liquid should be splashed on the hand. Only a very small quantity of the etching solution is required.

A pair of tweezers for handling the crystal is essential. They should be plastic tipped so that the etching solution does not come into contact with the metal. It was found convenient to coat the points of the tweezers with a solution of polystyrene in an organic solvent, allowing it to dry before use.

## Frequency Measurement

The crystal frequency should be measured as accurately as possible by means of a wave-meter or even an accurately calibrated receiver. The crystal has to be incorporated in a suitable oscillator circuit (such as that shown in Fig. 1) for frequency tests. The anode current is a measure of the crystal activity, being small with a crystal of high activity. The value of the anode current should also be found when the crystal has been removed from the circuit, as this is the value which indicates that no oscillation is taking place. The valve used was a 6AC7, but any sharp cut-off, high gm pentode such as the EF91, EF80, Z77. 6AM6, EF42 or EF54 would also have been suitable. No direct form of coupling was used between the oscillator circuit and the wave-meter, stray coupling being quite adequate.

The crystal should be removed from its holder by lifting it by the edges with tweezers, care being taken not to touch it with the fingers. If it is dirty or if it is accidentally touched with the fingers, it should be cleaned with chromic acid or with an organic solvent, as it is essential that the etching solution should come into contact with the whole of the crystal faces.

## Etching Rate

The rate of etching depends on the remperature, the strength of the etching solution, the crystal frequency and the type of crystal cut. A maximum
(Continued on page 1010)

# m"menecking Transistor Circuits manm 

## FINDING FAULTS WITHOUT INSTRUMENTS

By F. G. Roger

## Checking Transistor Receivers Without Instruments

IF a transistor receiver fails to work or to give normal results, stage by stage checking will probably be the quickest way of locating the fault. The tests described here are particularly intended for constructors who have no instruments, or perhaps only a simple meter for reading voltages, etc.

A typical transistor circuit, with check points, is shown in Fig. 1. Most of the tests described can be applied to any transistor receiver, though the number of stages, and exact circuit details, will vary.

## Newly-Bullt Sets

When the receiver has been constructed, but fails to work correctly, wiring should be carefully checked against the wiring plans or circuit. Any wrong connection, omitted lead, or short circuit can prevent the receiver working. Particular care should be taken to see that switch tags, or coil and transformer pins, are not shorted due to fragments of solder or projecting ends of wire. At the same time check that all joints are securely soldered.

If no fault is discovered, and components are new, the alignment of aerial, oscillator coil, and intermediate frequency stages may be so badly at fault that no signals can be received. If so,
this can be corrected as described later. In a superhet such as that in Fig. 1, wrong alignment can make the set quite dead, though no actual wiring or component fault is present.

Transistor connections, and the pin identification of all coils and transformers, should be checked. If no fault is found, stage by stage checks can be made as will be described.

When examining wiring, assure that resistor values are correct. Colour coding is usual, and a mistake can be made in reading the values. A slight change in value from that specified will not usually have a severe effect, but a large error (such as in reading the number of noughts) may prevent the set working or cause very poor reception.

## Resistance Checks

If a meter with "Ohms" scale is available, suspected resistors may be disconnected at one end, and measured. If a simple milliammeter is to hand, resistors can be checked with a 3 V dry battery. From Ohm's Law, current will equal 3 divided by the resistor value in ohms. The result is changed to mA by multiplying by 1,000 . For example, a. 1 k or $1,000 \Omega$ resistor in series with the meter and 3 V battery will pass 0.003 A , or 3 mA . If no meter at all is available, suspected resistors should be checked again, to discover if any error was made in reading the values.


Fig. 1.-A typical transistor circuit with test points.

If these tests do not reveal any fault, stage by stage checks can be made. These checks apply both to a newly constructed receiver which has never worked correctly, and to an existing receiver which has failed.

## Method of Testing

The valve receiver is tuned to a station and switched on. For I.F. outputs, the valve receiver speaker can be muted by turning the volume control to zero. When using audio outputs. the

## Stage By Stage Checks

The purpose of these is to discover which stage in the receiver is at fault. Investigations are then localised to the few condensers, resistors, connections, and other items in the defective stage.
If headphones are available, they can be used to trace the audio signal. A $0.1 \mu \mathrm{~F}$ or similar condenser should be included in one phone lead. Referring to Fig 1, the phones may be connected from (1) to chassis or earth line. If reception is good, the first three stages are working. Transfer one phone lead from (1) to (2). If reception ceases. the volume control or resistor and condenser between (1) and (2) are suspect. Remove the lead to (3). Signals should be considerably louder If not, the driver transistor or associated connections and parts are defective. Try the lead on (4) and (5). If there is no signal at either point that half of the driver transformer, or connections to it, should be tested. If the signal is still available at (4) and (5), check at (6) and (7). A further amplified signal should be heard. If not suspect the associated transistor.

Similar tests can easily be carried out with a TRF type receiver. When the point at which the fault is present is passed, reception will cease. Unsuspected faults can often be quickly located by this method.

For checking all stages, a source of audio and intermediate frequency signals will be extremely useful. Such signals may, of course, be obtained from a generator. But assuming that such an instrument is not available, an ordinary radio receiver may be pressed into service.

Fig. 2 shows stages in a typical battery portable or similar receiver. If a station is tuned in, an audio output will be available at point (1). Point (2) also gives an audio signal, at greater power. The usual volume control allows the signal strength to be adjusted.
Many receivers have a $465 \mathrm{kc} / \mathrm{s}$ I.F., so a modulated signal at about this frequency may be taken from points (3) or (4). A short piece of flex can be used as a test lead. For audio outputs, use points (1) or (2), and include a condenser of about $0.01 \mu \mathrm{~F}$ in series with the lead. For $465 \mathrm{kc} / \mathrm{s}$ outputs, use points (3) or (4), and wire a 100 pF or similar condenser in series with the test lead.
Mains receivers may also be used. with appropriate care, and reliable isolating condensers in the test lead. With a TRF receiver. no intermediate frequency signal will be available.


Fig. 2.-Obtaining audio and I.F. outputs from a receiver.
speaker in the valve receiver must be muted. This can be achieved by connecting an equivalent resistance in its place.

To check the transistor receiver speaker and output transformer, if necessary, apply the audio signal at (6) or (7). With the A.F. signal injected at (4) or (5) volume should be increased. If the lead is removed to (3) and reception ceases, the driver transformer, or connections to it, are defective.
Taking the lead to (2) should give increased volume. If not, the driver stage is faulty. With the lead taken to point (1) output should be adjustable by means of the volume control. If not, this component, or wiring to its tags, must be checked.

When working backwards from the output stage in this way, reception will cease when the point of the defect is passed. Also look for additional volume, as each additional stage is brought in, and reduce signal strength with the valve set volume control, to avoid overloading.
These tests will localise any defect in the audio stages of the receiver. Once this section has been cleared, or the defect found, attention can be given to the I.F. stages.

## Checking I.F. Amplifiers

An intermediate frequency output is now obtained from the valve receiver, as described. The test lead is taken to point (8). If results are good, the third I.F.T. and diode are working. If signals were obtained with audio injected at (1), but not
with the I.F. signal injected at (8) suspect the third J.F.T., or diode, or associated connections.

If results are obtained, transfer the lead to (9). Amplified signals should be heard. If not, suspect the second I.F. transistor, and associated resistors and wiring. If a signal is obtained, and valve receiver and transistor receiver both have a similar intermediate frequency, it should be possible to adjust the core of the third I.F.T. for maximum results.
The lead is then connected to point (10). If reception ceases, the second l.F.T., or connections


Various types and makes of tranisistor differ in appearance and three of the common types are shown above to facilitate identification.
to it, must be checked. When the lead is taken to (11), volume should increase. If not, suspect the first I.F. transistor, or associated components.

## AVC Action

As more stages are brought into use, it will be necessary to avoid overloading the transistor receiver. A very small serics condenser should be used in the test lead. This can be arranged by twisting two insulated wires together for an inch or so, and reducing the length twisted together if signals are too strong. When testing the I.F. Stages, keep the transistor receiver audio volume control at maximum, and reduce volume if necessary by keeping down the strength of the signal injected.

With the lead connected at (11), or merely placed near this wire, the second and third I.F.T.S can be aligned. The lead may be transferred to (12). If reception ceases, suspect the first I.F.T. or connections to it. If results are obtained with the lead at (12) but not at (13) then the oscillator coil collector winding it defective, or the coil wrongly wired.

If the two receivers have the same intermediate frequency, and one of the transistor set, I.F.T.s cannot be adjusted for best results, connections to this transformer should be checked. The injected signal must be kept at a very low level, as explained, or the automatic volume control circuit of the receiver will make adjustment difficult.

If an intermediate frequency amplifier is suspected, it may be temporarily eliminated, with its transformer, by connecting point (11) to point (9). in the case of the first I.F. stage. Or the first I.F.T. can be eliminated by connecting point (10) to point (12). and so on. Any unused transistor should be unsoldered. One I.F. stage transistor can be compared with the other by using only one I.F. stage in this way, and trying the transistors in this stage in turn.
When the defective stage has been located, check wiring and resistor values carefully. If all com-
ponents are in order, and connections are correct, the associated transistor must be suspected.

## F.C. Stage

When good results are obtained with a weak signal injected at (13), but the receiver will not operate, the F.C. stage is probably defective. If the defect shows up on one waveband only, this would have immediately made the associated coil windings, switching, and wiring suspect. These items should in any case be checked, when a fault has been localised to the F.C. stage.
To eliminate the aerial circuit, iead (14) can be disconnected from the switch, and taken to an exterior tuned circuit. This can be any coil and variable condenser tunable to the medium or long wave band. The coil is returned to the earth line, or base feed network. If good reception is then possible, the acrial circuit is probably faulty.
Alternatively, an aerial may be connected to point (14). If reception is possible. the acrial circuit may be faulty, or trimming badly in error. But if neither of these tests allow any signal to be received, after the checks previously described, then the F.C. transistor is probably not oscillating. Oscillator coil connections should then be checked.

When reception is good with an acrial, but ceases when this is disconnected, lack of alignment alone may be responsible. If so, adjustment of the oscillator coil core, or aerial coils, or trimming, will remove the trouble. Remember to adjust cores at'a high wavelength on each band, and trimmers at a low wavelength. Alignment is correct when no further improvement in volume is possibie. As reception is improved by early adjustments, select weak stations for subsequent adjustments, to prevent the AVC action already mentioned.

## Other Tests

If a new battery is not being employed, it should be checked with a meter. The correct polarity is absolutely essential. Switch, and positive and negative circuit continuity, can be checked by working along point by point from the battery.

A meter and small battery can also be used to check continuity of aerial. oscillator coil, I.F. transformer, and other windings. Disconnect one end of suspected windings first.

In must cases only a few of the tests described will be required. For example, if A.F. is applied as explained, working back from the speaker, a defect in the A.F. amplifier may be at once located. Similarly, if the A.F. amplificr is cleared. applying an I.F. signal to points (8) through to (13) will quickly show where the fault is.
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# A STERED AMPLIFIEIR 

8W WITH SINGLE-ENDED OUTPUT CIRCUITS

By R. Hindle

## (Continued from page 902 of the Fehruary issue)

7HE main circuit is fully conventional. R4 C1 and R12, C4 serve to decouple the first audio stages. The gain controls (VR1, VR2) are placed between the two audio stages where the signal is of the order of 1 V . It will be seen that the H.T. switch connects the centre-tap of the $300-0-300 \mathrm{~V}$ winding to earth via the indicator lamp L1 which, when the switch is in the "on" position, carries the H.T. current of the whole amplifier, or more correctly the charging current to the capacitor C8 which is of pulse form. Normal current lights this lamp quite adequately, the initial surge on switching on being evidenced by the brightness of the bulb. Excessive current will burn out this lamp, however, so that it acts as a fuse. The transformer specitied has two $6 \cdot 3 \mathrm{~V}$ centre-tapped windings in addition to the winding for the rectifier heater and these are used one for each amplifier line. It is not essential to feed these from separate windings if a transformer with only one such winding is available. The "mains on" indicator lamp may be across one half of one of these windings, this being found to give quite adequate light with the lamps used. Centre taps go to earth, of course, and both sides of the heaters of the valves are floating.

The gain of this amplifier even without a preamplifier is likely to be adequate for the popular type of stereo cartridge of the crystal or ceramic type and the constructor may like to try it out coupled directly into this amplifier pending the production of
the pre-amplifier. In this case it should be borne in mind that the crystal pick-up has to be correctly loaded in order to produce the correct frequency characteristic. The appropriate resistance is quoted by the makers and this value should be substituted for R1 and R9.

## Construction

The amplifier is constructed on an aluminium chassis $12 \mathrm{in} . \times 6 \mathrm{in} . \times 3 \mathrm{in}$. The drilling diagram was given in Fig. 3, which shows the sides of the chassis as if they were opened up. The layout shown is as on
 support the volume control.


Fig. 5.-The under chassis wiring diagram.


The amplifier with all the wiring completed.
the outside of the chassis. The actual components to be used should be measured of course before drilling in case they are found to vary somewhat from the sizes shown. No mounting holes are drawn for valve holders and electrolytic condenser clip. The best way is to drill the hole as shown, drop in the component, turn it until the pins are in the relative positions shown on the wiring diagram (Fig. 5) remembering that this is the underneath of the chassis, and then mark the positions for the fixing bolts.
The mains and output transformers are of the shrouded type with soldering pins projecting underneath. Slots have to be cut in the chassis to allow passage for these, the method being to drill holes at each end of the slot position, then cutting away the metal between the holes. The positions for the possible additional valves are shown but not the holes themselves. The position of the screen is shown in Fig. 3, and the dimensions of the actual screen are in Fig. 4, from which the positions of the fixing screws on the main chassis can be taken.

## Wiring

When the components are mounted, wiring should begin by feeding the heater supply with twin plastic flex, running this close to the chassis. A soldering tag under one of the securing boits. of each valve holder is used for all earths for the circuit associated with that valve. The only exception is the earth for the input socket which goes to a tag on one of the bolts holding this socket. All resistors are selfsupporting on the valve-holders and are put as close as possible to the holder. The input lead from the socket to the valve is a short piece of coaxial cable of which the screen is earthed at the input socket end

If two signal lamps are to be fitted it will be found that the light from one bulb will shine through the window of the other, so obscuring the meaning of the lights, unless an aluminium shield is shaped and fitted between the lamps (or, of course, the enclosed type of signal lamp could be used).

Note that the output transformers can be used for either $15 \Omega$ or $3 \Omega$ speakers according to the connections made to the secondaries. The wiring diagram shows the two secondary windings of each transformer connected in parallel for $3 \Omega$ working. If $15 \Omega$ work-
the chassis to allow easy connection.

## Surrey's Most Modern Factory

The second stage in the erection of the new factory at Chessington for the Solartron Laboratory Instruments, Lid. the electronic instrument manufacturing company of the Solartron Electronic Group, Farnborough, Hants., is now under way.
The administrative block of 25.000 sq. ft , which presently houses the instrument sales and the international division of Solartron, was occupied early in 1960. Also in this block are the instruments servicing and works training sections, together with some production facilities.
The second section of 50,000 sq.ft being built will be completed in October, 1961, and the third stage-a further expansion of 50.000 sq.ft, for which the Chessington site has available spacewill be completed in three years' time. The final factory will be one of the most modern in Britain for the manufacture of electronic instruments and will employ more than 1.000 pcople.
The training centre, which is now incorporated in the new buildings, is claimed to be the most advanced for workers in electronics. A special study has been made of training methods. and the result is that operators are working confidently at full production at their benches after an initial training period of six weeks. The training continues from time to time in intensified courses for two years for the creation of an expert operator. In fact, training is integrated with production somewhat as is a University "sandwich" course.
The Sola tron Electronic Group is a member of the Firth Cleveland Group of Companies. under the chairmanship of Mr. Charles Hayward, and currently Solartron's sales of electronic test instruments alone exceed $£ 2,000,000$ per annum, and it is planned to double this amount over the next five years.
The Solartron Group Director responsible for the Chessington factory is Mr. E. R. Ponsford, Managing Director of Solartron Laboratory Instruments, founded 13 years ago as the first Company of the Group.

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## Aerial Design

0NE important feature arising out of correspondence from some of my readers was regarding the choice of aerial suitable for short-wave section. In the old days one was content to erect a large horizontal aerial system, usually running the full length of a garden, with a large mast at each end and the lead to the set was taken from one end or from the centre. This type of aerial (an inverted ' $L$ ' or a ' $T$ ') was a good all-round arrangement, and many newcomers are inclined to eopy this arrangement and in spite of the all-important feature of height and insulation, and the use of the multi-stranded wire, only ohtain very indifferent results. 1 am asked why this is. The main reason is that nowaditys the majority of high-powered short-wave stations use a very high metallic mast which itself is the radiator or aerial. Under these conditions, therefore the signal is vertically polarised. As most readers know. it is desirable that the recciving acrial and the transmitting aerial should be in the same planecompare the television aerial or the VHF aerial. for instance. The types of aerial used earlier and referred to ahove, are horizonlal types and must naturally, therefore, he productive of inferior results. I have found that a very gond aerial for all-round short-wave reception (that is. for a receiver which may he used on a large number of bands and not tied down specifically to one amateur band) consists of a vertical wire. The arrangement which I have tried out and have indeed recommended to a number of amateurs. is to fit a length of stout battening to the soffit carrying the gutter and to the end of the batten attach an insulator. One end of a length of wire should be attached to the insulator which should be at least 3 ft from the walls of the house. The lower end of the wire should be anchored to a similar arrangement of batten and insulator fixed to the lower part of the house so that the wire is about 15 ft long. The lead to the receiver is taken from the lower end. In the majority of cases this arrangement will be found productive of very good all-round results except in some cases, where the house acts as a screen. Except on the very short wavelengths, this should not have a very marked effect. and the only ill-effects which have been noted were in the case of very heavy rain-storms when the wall immediately behind the aerial became soaked, and signals did then drop off. If you have not tried such an arrangement, and are experiencing indifferent results in your short-wave
working, perhaps you would try this schème and let me know how you get on. I should like to stress that letters from my readers are always appreciated, especially where some scheme has been tried out and I will always be pleased to pass on the results of tests which may benefit other readers.

## Rain Effects

Whilst on the subject of aerials I should like to mention a peculiar effect which was reported to me by an acquaintance. He was listening to the short-waves one evening and suddenly a ticking noise started on the band on which he was listening and he thought that it was an amateur starting up his transmitter, or a new beacon which he had not previously heard. He found, however, that the noise extended over the entire band and was not tunable. For a time he put it down to some form of electrical interference, hut as it continued he began to wonder whether something was wrong with his equipment and so started to take particular notice. The tiching had an almost clockwork regularity, and was of very constant strength. Suddenly the regularity fell off and longer intervals oecuried between the individual "clicks", and after a very short time they ceased. He was left wondering what had been the cause "of the trouble and then continued his "scarching". After a few minutes the noise started again with the constant regularity, and fortunately at that moment the wind apparently changed and he heard rain beating against his window. This gave him an idea and he looked out at his aerial and from the insulator nearest his house he sàw a constant steady drip of rain. The water was running down the aerial, over the insulator to its lowest point, and then dripping from there. There was obviously snme electrostatic effect taking place, and when he pushed up a length of wood to take away the drips the noise ceased, as it did when it stopped raining. The insulator, incidentally was one of the ex-government Pyrex long components and was not fractured-but it was coated with a very heavy deposit of soot. He gave it a good scrub with warm water and a detergent and at the next heavy shower it gave no trouble at all. Perhaps. therefore, if your aerial equipment has been up for some long time it might be worth while lowering it and giving it a clean up; a heavy sooty deposit could cause signal losses.

## Heat Sinks

To obtain maximum efficiency from heat sinks used for power transistors, paint them dull black. Black is a much better radiator of heat. so that a small hlack heat sink may be much better than a larger shiny one. The principle ciln also he used for ordinary transistors. If a transistor is going to be near something warm. paint it with a shiny aluminium paint. If you are worried about the transistor itself heating, paint it dull black so it can easily radiate its excess heat.

# A Versatile TAPE RECORDER 

 author, the front panel should be made llin. wide. If the recordlevel indicator is to be built into the recorder, the front panel should be drilled as shown in Fig 5 but if not the panel should be drilled as in Fig. 4.For the record-level indicator (whether built into the recorder or not) a small moving-coil milliammeter is required. Any value of full-scale deflection up to a maximum of 5 mA can be used. An old thermo-couple R.F. meter with the thermo-couple removed should be ideal as the scale readings are un-important-a linear scale is not required, any calibration will do. The diameter of the meter-case should be about 2 in . although up to $2 \frac{1}{2}$ in. can be fitted on the panel. A hole to suit the meter should be cut in the panel including the three or four fixing holes required. If the depth of the meter-case is greater than $1 \frac{1}{1} \mathrm{in}$ the meter will not fit flush with the front of the panel as it will touch the second and third valves. This small difficulty can be overcome by using small stand-off washers to keep the front of the meter about $\frac{1}{4} \mathrm{in}$. from the panel.

## Drllling

When drilling the front panel it is advisable to make sure that the lower holes coincide, as near as possible, with the similar ones on the front wall of
(Continued from page 899 of the February issue)


## The Function Switch

As stated before, the function switch is very important. A good quality well insulated 3-pole, 3 -way, 2 -bank switch is required. It must be of standard size and not a miniature. This is important as the stray capacitance between tags on a miniature switch is high and such strays must be kept low in order to avoid instability and unwanted feedback.

First of all, the switch is dismantled, except for the ball-bearing mechanism, and all parts cieaned, especially the two wafers. These two wafers will, for convenience, be named Bank 1 and Bank 2. A small piece of Mumetal, radiometal or similar high-permeability metal is now needed for a screen between the two banks. Ordinary tin-plate would do the job. but not as efficiently as the Mumetal. An old cathode ray tube screen would be a suitable source of high-permeability alloy. The size of the screen is $3 \frac{1}{4} \mathrm{in}$. long by $2 \frac{1}{4} \mathrm{in}$. wide and it should be cut and drilled as shown in Fig. 6 although the measurements are nominal and depend on the size of the switch.

Holes $P$ and $Q$ should be drilled to suit the two fixing bolts of the switch. Two small
the chassis. The panel and chassis should now be cleaned with a cloth to remine all small metal particles and dust, before being put to one side.
grommet-holes are needed as shown. The switch is now built up as illustrated in Fig. 7. More details are given in Fig. 11. After bolting the switch together,


Holes marked 'A' drilled to suit cosxial sockets
Fig. 5 (above)-Alternative front panel drilling details when a record-level meter is to be included.
Fig. 6 (below)-Dimensions of the Mumetal screen for the function switch.
Holes marked $P$ and $Q$ are switch fixing holes

it can be partially wired. The connection from SA3 to SC3 (Fig. 2, page 898 last month) goes through the lower grommet-hole in the screen, whilst the two earthed sections on Bank 1 and Bank 2 are connected via the upper grommet-hole. It should be noted that the two banks are back-toback and that the tags of Bank 2 face the balllocating mechanism. In all cases SA, SB and SC, SD indicate the moving contact and the numbers 1,2 and 3 the fixed contact being touched by the wiper. Although the switch, when described, seems a little complicated, its function and building are easy to see once one actually handles the switch and can inspect the various contacts.

## Wiring the Amplifier

The most difficult part has now been completed and the actual wirting of the amplifier can be commenced. Two tag-boards are needed. one 16-way, and one 5 -way. Roth should have two conlacts per tag and about $1 \frac{1}{4} \mathrm{in}$. between the inner tags. The larger tag-hoard is wired first: a complete wiring diagram is given in Fig, 10. It is best to wire in all connecting links before any actual components. Short leads of tinned-copper wire should be fixed to the lower, outer contacts, as shown,
amplifier. Ordinary carbon resistors are in the be noisy in operation. If such resistors to used in a high-gain stage, any small fluctuation in the current through them is amplified and produces quite a high background noise level. Thus for resistors R1. R2. R 3 and R4, good quality cracked-carbon, high-stability components should be used. This type of resistor need not be very expensive-in the prototype amplifier, all $\frac{1}{2} \mathrm{~W}$ resistors were 10 per cent tolerance high stability components.

C 2 and C9 provide decoupling in the amplifier in conjunction with R16 and R5. These two capacitors are wire-ended tubular electrolytics, as are C1, C12. C19 and C22. Good-quality paper tubular capacitors must be used for C3, C10 and C14 as any leak in C3 and C10 will upset the biasing of valves 2 and 3 , whilst a leak in C14 would pass a direct-current through the record/ playback head and magnetise it. The equaliser circuit can now he built on to a small 5 -way tagboard. The equaliser described uses a network of resistors and capacitors, but no inductors, as is more usual. In the recorder it was desired to use the fastest speed of $7 \frac{1}{2} \mathrm{in}$. per second for recording concerts and the speed of $3 \frac{3}{4} \mathrm{in}$. per second for ordinary work. Although a speed of $1 \frac{7}{8} \mathrm{in}$. per second is available on the tape-deck, it was decided to use this speed for private work only,


Fig. 7.-The assembly of the switch, including the Mumetal screen.

Which did not require good quality at high volume. Recordings made at $1 \frac{7}{3}$ in. per second are good, but only if played back quietly. High volume tends to make the noise-level in very quiet passages too much of a distraction.

## Treble Boost

It will be seen from the frequency response
speeds although it must be stated that it is a compromise, providing theoretically correct equalisation at 6 in. per second. R1I controls the amount of bass equalisation, and provides perfectly correct results at both speeds. being fully variable. Fig. 9 shows the approximate record/playback characteristics. The layout of the equaliser circuit and its tagboard are shown in Fig. 10. An alternative equaliser

graph that a great deal of treble boost is required above $1500 \mathrm{c} / \mathrm{s}$. In many recorders, it is the practice to apply trebie equalisation during recording and hass equalisation during playback, but for this recorder, treble boost is applied both during record and playbach, whilst bass correction can be made on record or playback, or both. However, the magnetic tape is easily overloaded if too much bass equalisation is made during recording and only cnough bass should be used to counteract any lack of it in the recording signal.

The circuit used for both $7 \frac{1}{\mathrm{i}} \mathrm{in}$. and $3 \frac{3}{3} \mathrm{in}$. per second was shown in Fig. 2 (page 898, Feb. issue). This circuit gives ample treble equalisation at both

Fig. 9 (right)-Equalisation curves.

Fig. 8.-Drilling plan of the main chass/s-holes for tag-strips, etc., have not been included.



Fig. 10.-The main and equaliser circuit tag-boards.

circuit is shown in Fig. 12 with components from Fig. 2 to give the best equalisation at the three speeds. Either of the alternative equalisers can be buith on to a. 4 or 5-way tag-board. For C ba trial and error would probably give the best results, using a 300 pF variable trimmer. For either type
of equaliser, the tag-board should be mounted on to the chassis using a small bracket, and details are given uflıg. 13. The amplifier proper can now be built. Grommets and valve holders are put in
(Continued on 1021)

# THE P.W. SIGNAL Masmin win GENERATOR 

MANY constructors find they need a signal generator, but few eventually possess such a unit-their cost is generally too great. Although most enthusiasts cannot afford to buy a signal generator, they would nevertheless like to build one but are hampered by lack of plans. This new series has been especially prepared for the home constructor and will describe in detail the construction of a wide. range signal generator.


By E. V. King

by soldering as shown in Fig. 3. Make sure that tags 2 to 6 inclusive do not short to earth.
Grommets should be fitted in the holes A, B, and C. The switch $\mathrm{S}_{1}$ is fitted with the slot in the threaded portion uppermost, the switch is then "on" when pressed down. Leave plenty of thread showing outside the chassis and tighten the nut. For the position of S 1 see Fig. 3.
The warning light is fitted from the inside using two nuts and bolts.

## Component Positions-Inside the Chassis

Fig. 3 shows the position of the transformer. It is bolted tightly to the tin plate and should have a small strip of mild steel fitted on top of the chassis. This strip is shown shaded in Fig. 2 and the general method of mounting in Fig. 6 . If this is not incorporated the unit will work well, but will give rather more hum than necessary.
A small tag strip with two isolated
Fig. 1 (left)-The circuit of the power unit.

The chassis used for all the units are the same, basically a baking tin 6in. $x 4 i n . \times 2 \frac{1}{2}$. high, and are easily obtainable. The bottom of the tin, after drilling the necessary holes for valve base, condenser and grommets, is soldered in position. The seam down one side of the tin is also soldered on both sides to give additional strength. There is no reason why readers should not make their own chassis or use aluminium instead of tin plate. If aluminium is used every earth connection to the chassis must be made via a solder tag and tight
nut and bolt.

## Component Positions-Outside the Chassis

Reference to Fig. 2 will show where to dill for $V_{1}$ and C1/C2. The hole for V1 is $1 \frac{1}{8} \mathrm{in}$. in diameter, and can be made with an ordinary carpenter's bit, chassis cutter, tank cutter or using small tin snips. The condenser hole will probably have to be a little larger. The valveholder may be holted or soldered in position and the condenser clip bolted up tightly, Verify that the valveholder is an International Octal, as damage to the valve pins will be caused if the wrong one is used. Fit the valveholder so that the central "keyway" is in the position shown in Fig.?
The tag strip is attached to the rear of the unit
tags is soldered to the side of the chassis near the
primary side of the transformer. primary side of the transformer. (Fig. 3.)

## Alternative Component Positions

If an upright-mounting transformer is used, it will have to be fitted on the top of the chassis. The component positions remain the same with the exception of the following points.
The condensers C3 to C6, which are wired in, will be inside and not outside the chassis. The transformer itsclf will be outside in the position shown in Fig. 8, and grommet holes will be required at E, D, and F. Another 6 -tag strip (with 4 tags isolated) is required and is soldered in place inside the chassis about tin. from the top, as in Fig. 9 . Make sure that no tags are earthed other than those at the extreme ends (i.c. 1 and 7 in Fig. 9). All other components are fitted during the
wiring process.

## Wiring the Unit

It will pay the beginner dividends in both time and patience if he develops the habit from the start of checking, double checking and testing at every stage of his work. This will be explained in cletail. Refer to Figs i. 2 and 3 as you proceerl. Take "C "mains lead through the grommeted hole "C" and tie a hnot in it. Connect the
black wire to one unearthed tag of the small tag strip. Also solder one primary lead from the transformer to the same tag. Check that wall soldered joints are well made, strong, and with no sharp edges. Now solder the red mains lead to any tag on S1. Solder the other primary lead to the unused tag of S1 If a three-core cable is being used, solder the green lead to the chassis. The chassis may be separately earthed if desired, but it must be earthed in some manner.

## Testing the Transformer Primary

Make sure that the leads on the secondary side of the transformer are not touching anything. Plug in and switch on. A hum should be heard. If no hum is heard check the power supply, mains lead, Sl and the primary wiring.

If a loud hum is heard check that the transtormer is correctly mounted with the steel strip above, and that the secondary wires are not shorting.

## Wiring the Heater Circuit

Scrape and tin the thick $6-3 \mathrm{~V}$ on the secondary side of the transformer; solder one to the chassis (use a large very hot iron for this joint) and connect the other to one side of the warning lamp holder. Connect the other tag of the holder to earth (always use a very hot iron for chassis connections). Screw in a low wattage, 6 V lamp. Plug into the mains and switch on. Ensure that the lanp lights. If it does not and the transformer appears to be operating correctly check the lamp, the lampholder and the heater


$\begin{array}{cc}\text { Warning Lamp Meial stiffening bar Taps land } 7 \text { can be soldered } \\ \text { for mounting Tr } 1 & \text { or bolted to metal chassis }\end{array}$
Fig. 2.-The ahove chassis view of the unit.
Refer to the valveholder and looking under it make sure you have the lags numbered correctly, sometimes this is not always easy. Be quite sure about this. Solder a wire from chassis to pins 1 and 2. Also solder a wire from pin 7 to the uncarthed lag of the lampholder. Plug in the 6×5 valve and chech that when connected to the mains and switched on, the heater is alight. If it is not, check the valve pin numbers and the value itself. Take an insulated wire from the unearthed tag of the lampholder through the grommet at B and on to tag ? above the chassis (see Fig. 2). This is the main external heater connection.
Take the lamp out of the holder and hold it between tag 2 (Fig. 2) and earth (chassis). Switch on. The lamp should light. If not. the last wire added has been wrongly connecied or tag 2 is shorting to earth. Put the lamp back in the correct holder and test again.

## Warning

When carrying out this procedure of building and lesting. never be temnted to work with the unit plugged in. In any case. when working in a garage or other location with a concrete floor it is wise to have a thich rubber mat on the ground.

## Wiring the H.T. Circuit

Temporarily connect the two red secondary leads of the transformer to a small wattage 15 or 25 W) main lamp or̀ mains neon lam. Switch on and chech that the secondary H.T. is all right. If this lamp does not light, hut the one in the holder does eheck that the lamp itself is not to blame. that the holder is correctly wired and fitted and that the wires ars not shorting.

Disconnect the mains lamp, and connect one secondary wire to the chassis. Connect the other to pins 3 and 5 of the valveholder. If a voltmeter is availatale connect it on the 250 V range (D.C.) to the chassis (negative) and ain S of the valveholder (nositive). Switch on, and the meter should read ahout 100 to 150 V D.C. The needle may quiver a little as the supply is unsmonthed. If no reading is obtained the valve is suspect.

Connect pin 8 through a resistor (R1) to one positive tag of the large electrolytic smoother $\mathrm{C} 1 / \mathrm{C} 2$. You will need to examine this carefully as they vary somewhat in their coding. One tag is the negative pole, usually either black or colourless. Two other tags are the Cl and $C 2$ positive tags and are usually, but not always, yellow and red. Sometimes the positive tags are red and colourless, so care is needed. Details are printed on the side of the can. if the instructions tell you that one positive connection is for the maximum ripple, that is the one to use for the connection to R1. Connect the negative tag to the chassis.

## Testing the Initial H.T. Supply

Connect the voltmeter on the same range as before, but with the positive lead on to the junction of R1 and C1. Switch on, and the D.C. voltage should be about 250 with no vibration of the needle. If not, verify that R1 is a good component and of the correct value, and that Cl is correctly connected.

## Wiring the Smoothing and Decoupling Circuits

Connect R2 so that it stands in air supported on the tags of the electrolytic condenser $\mathrm{Cl} / \mathrm{C} 2$. Connect the voltmeter as before, but with the positive lead to the junction just made between R2 and C2. Switch on and the voltage should be about 250. Switch off. Leave for 30 seconds. Short, with an insulated screwdriver, R2 and earth (chassis) when a spark should jump, showing that the condenser $\mathrm{C} 1 / \mathrm{C} 2$ is working properly, and holding a charge. Henceforth, whenever you are working on the unit after switching it off always discharge the condensers twice at 15 -second intervals before recommencing work.

Connect an insulated wire to the junction of R2 and C2, lead it through the grommet " $A$ " and bare the end for $2 \frac{1}{2}$ in. to allow connection to resistors R3 to R6.
With reference to Fig. 2, cut the leads to R3, 4, 5 and 6 so as to leave about $\frac{3}{3}$ in. lengths on each. Solder them as shown between tags 3, 4, 5 and 6 and the bared lead. Bend them more or less upright. Switch on the supply and connect each tag ( $3,4,5$ and 6) to the voltmeter positive in turn. The voltage reading will depend on the voltmeter used. Expect a drop of the following voltages from the reading previously taken at the junction of R 2 and C 2 :

\[

\]

If one tag gives a different voltage from the others then the associated resistor may have been wrongly chosen. If any tag gives zero reading


Fig. 3.-Underchassis wiring and layout.
it is earthing to chassis or the resistor is completely open circuit. Check by substitution.

Now solder in the four decoupling capacitors (C3 to C6). These need not all be the same type, but the constructor must again make sure of the polarity of the connections. Solder them as shown in Fig. 2, between the tags and the chassis. Negative ends going to chassis always.

Now check the voltage at each tag again and although they may vary somewhat ( 10 V or so) from one to another, all should give over 200 V . Switch off and quickly short each in turn to chassis with an insulated screwdriver. Each should give a good sharp spark and "crackle". Remember to discharge twice at junction R2/C2 as previously explained.

## Specification of the Power Unit

The unit will operate from 200 to 240 V A.C. mains and gives four outputs, almost independent of each other, at about 220 V (low current). A 6.3 A.C. filament supply is also available. The unit may be used for other purposes than that in mind, and readers may take H.T. at 250 V at up to $50 \mathrm{~m} A$ from the junction of $\mathrm{R} 2 / \mathrm{C} 2$ for driving amplifiers, tuners, etc. Being small, the unit is very useful in the radio laboratory. In this unit the various outputs will be used as follows:

Tags 1 and 7, earth (chassis), this is H.T.- and one side of L.T.
Tag 3, R.F. oscillator and cathode follower;
Tag 4, L.F. $400 \mathrm{c} / \mathrm{s}$ oscillator for modulation;
Tag 5, L.F. variable Wien bridge oscillator (A.F.) and crystal calibration unit:

Tag 6. amplifier for variable A.F. oscillator;
Tag 2, heater lead to all units.

## Using Upright and Larger Transformers

The type of transformer used is not critical so long as the ratings are those specified in the components list. The placing and mounting of the transformer will depend on its construction but
(Continued on page 998)

# STYLUS WEAR TIMER 

## AN EFFICIENT MEANS OF KEEPING A RECORD OF STYLI PLAYING TIME

By F. Churchill

$R^{2}$
ECORD manufacturers point out that styli may be worn and in need of replacement long before this is made apparent by increased surface noise; with consequent damage to valuable records. The recommended life of a diamond tip is about 2.000 hours, and that of a sapphire about 40 hours. There are probably some methodical enthusiasts who keep a careful note of playing times. but 1 suspect that most of us quickly lose the habit. By using the simple little instrument described below, it is possible to know at any time. exactly how long a particular stylus has been in use.


Fig. 1.-The method of coupling the motor to the cyclometer.

## Components Required

The parts required are, a small "shaded-pole" synchronous motor fitted with reduction gearing giving hours instead of miles, a cyclometer or small counter, a piece of $\frac{1}{16} \mathrm{in}$. sheet brass- $2 \frac{3}{4} \mathrm{in}$. $x \frac{3}{3} \mathrm{in}$.-and a few 4B.A. nuts and bolts. The motor complete with suitable g-aring can be obtained at moderate cost from suppliers advertising surplus equipment. (Note. an ordinary electric clock will not do. because it is not self-starting.) Whilst it is possible to get the counter from a similar source, if one buys a new cyclometer, this will be supplied complete with an adjustable bracket, and thus make it unnecessary to construct one from brass sheet.

## Transmitting the Drive

The method of transmitting the drive from the final spindle to the counter will depend on the type of spindle fitted to the latter instrument. A true cyclometer is fitted with a "star" wheel, and this type can be driven by a bent arm fixed to the gear spindle and engaging in one of the teeth of the "star" wheel. The author's cyclometer has a spindie which is slotted at one end: consequently it was only necessary to file two flats on the end of the gearing spindle. which, when engaged with the slot, formed a positive drive.

Many of these small motors with gear-train have a spring-loaded clutch on the final spindle, to facilitate altering the hards when used as a clock. It is essential that this clutch is made inoperable, so that the drive is direct. If this is not done, there will be a tendency for the drive to slip whenever the load increases, e.g. when the figures change from 99 to 100 , etc.

It is advantageous to fit a length of thin twinflex to the existing motor leads so that the unit can be placed in any convenient position; taking care to keep well away from pick-up leads and the pre-amp. to avoid inducing hum. These leads are connected to the terminal block on the gramophone motor. so that the counter only operates when the gramophone motor is actually switched on.

As fractions of an hour are not required, it will avoid confusion if the extreme right-hand set of figures are obscured with a piece of opaque Sellotape, or by some other method.


Fig. 2.-An alternative method of coupling the motor to the cyclometer.


Fig. 3.-Dimensions of a saitable mounting bracket.

## CORRECTION

In an advertisement for Stern Radio Lid., which appeared in the February issue of Practical Wireless. the price given for "A Complete Kit of Parts for Stereo Pre-amplifiers" should have read £12 10s. Od. and not £il 10s. Od. as printed

# A valve voltmeter 

THIS INSTRUMENT GIVES READINGS FROM I TO 1000V.

By J. B. Dance

IF an ordinary voltmeter is connected across two points in a circuit, it takes a certain amount of current from the circuit. The voltage across the two points will normally decrease when a current is being taken owing to the internal resistance of the circuit, and therefore the true voltage between the points when the voltmeter has been removed is not known. This error is often negligible, (e.g. when measuring power supply voltages) but it always exists and may be very large in high-resistance circuits. The error can be minimised by using a voltmeter which has a high internal resistance in comparison with the internal resistance of the circuit being measured.


Fig. 1.-This circuit may be used over nine voliage ranges.

## Moving Iron Voltmeters

Moving iron voltmeters invariably have a low internal resistance. An expensive moving coil $0-50$ microammeter has an internal resistance of $20,000 \Omega / \mathrm{V}$ at full scale deflection when it is used as a voltmeter; even this is not high enough for some purposes, but all cheap meters have an internal resistance much lower than this.

The four main types of simple apparatus which can be used to measure voltages without taking any appreciable current are:-

1. neon bulb circuits,
2. electrostatic voltmeters,
3. potentiometer circuits,
4. valve voltmeters (including methods using cathode ray tubes).


Fig. 2.-The test-probe for D.C. instruments.
Methods 1 and 2 cannot be used to measure voltages of values less than about 100 V . Method 3, when used in skilled hands, is excellent, especially for low voltages, but method 4 is probably the most convenient method when quick direct readings over a large varicty of ranges is required.
The circuit of a simple valve voltmeter is shown in Fig. 1. This has nine D.C. ranges shown in Table 1 (page 998), with full scale deflections varying from 1 V to $1,000 \mathrm{~V}$. It can also be used to measure audio and radio frequency voltages below 50 V . The D.C. input resistance is greater than 20 M and is constant from range to range. On the one voltage range, at full scale deflection, the current taken by the instrument is about $50 / 1000$ of a microamp $(20 \mathrm{M} / \mathrm{V})$ whilst on the 1000 V range the current taken is $50 \mu \mathrm{~A}$ at full scale deflection ( $20 \mathrm{k} / \mathrm{V}$ ).

## D.C. Probe

The probe shown in Fig. 2 is used for D.C. measurements. It is connected to the main instrument by coaxial cable. The body of the probe consists of a "Tuffnol" tube with end pieces which are glued on to the tube after the resistor ( R 16 ) and the stand-off insulator have been fitted inside the probe. The wire end of the resistor is used as one contact, the other contact being made by means of a wire connected to the outside of the coaxial cable. Coaxial connection to the main instrument is used in order to prevent hum voltage pick-up in the high resistance circuits.

## Components

All of the resistors, R2 to R10 should, if possible, be high-stability components and it is desirable that they should have a tolerance of $\pm 1$ per cent. If great accuracy is not required, however, $\pm 5$ per cent resistors could be used. R1 and the D.C. probe resistor, R16, should be high-stability components, but their tolerance is not important. If high-stability resistors are not available, it should be possible to obtain reasonably accurate readings by using the ordinary carbon type. The unusual resistor values can be made by using two or more resistors in series or parallel combinations.

A ceramic wafer switch was used for $S 1$, so that the unwanted leakage current would be kept to a minimum.

Cl should be selected for high leakage resistance. A good quality component should have a leakage resistance of over $2,000 \mathrm{M}$.

The valve used may be a $12 \mathrm{AU7}$ (B9A base) o a 6SN7 (octal), but almost any small double triode would be satisfactory providing that the constructor is willing to adjust the resistor values to suit the valve chosen.

The double-pole double-throw switch $\mathrm{S} 2(\mathrm{a})$ and $\mathrm{S} 2(\mathrm{~b})$, in the valve anode circuits, is provided so that the meter deflection can be arranged to be in a forward direction, no matter whether the voltage applied to the grid of $V 1(\mathrm{a})$ is positive or negative. If a centre reading microammeter had been available the switch S2 would not have been required. The meter was calibrated from 0 to $500 \mu \mathrm{~A}$ and it was therefore easy to read the voltages being measured directly in volts for all of the ranges shown in Table 1.

## Principle of Operation

The voltage from the probe is fed into the potential divider resistors R1 to R10. The total resistance of the potential divider (including the resistor in the probe, R16) is constant whatever the position of S1. The range is selected so that the switch S1 taps off a voltage from the potential divider which the valve circuit is capable of measuring (a fraction of a volt). R11 and Cl filter out any alternating voltages before the input reaches the grid of the valve. VR2 is adjusted before use so that the meter reading is zero when no input is being applied.

Let us suppose that a small positive voltage is applied to the grid of Vl(a). The anode current of this valve will then increase, making the anode voltage more negative with respect to the chassis. At the same time both cathodes become more positive owing to the increased current through the common cathode resistor, R14, and V1(a). This is equivalent to a negative bias on the grid of V1(b) and the current through this valve therefore decreases and its anode voltage increases. A current will therefore flow from the anode of V1(b) through the meter to the anode of V1(a). If a negative voltage is applied, the current will flow in the reverse direction.

## D.C. Callbration

The instrument should be used to measure any accurately known voltage which ought to give a full scale deflection on any particular range, e.g. 100 V . The value of R1 should then be adjusted until a value is found which gives almost exactly a full scale deflection on the correct range. The value of this resistor will generally be of the order of 10 M . Various high-stability resistors should be tried for R1, until a value which is very nearly correct is found. The final adjustment is then made by VR1. The value of this potentiometer should be kept reasonably small in order to preserve stability.

When one range has been calibrated, all of the others should be correct providing that the tolerance of the resistors R2 to R10 is within the limits of accuracy required.

## Simplifiod Version

The number of high-stability resistors required can be reduced by constructing an instrument having fewer ranges. The ranges can be selected to meet the

> COMPONENTS LIST
> RI-10M (see text), high stability.
> R2-8M $\frac{1}{2} \mathbf{W}$ high stability.
> R3-1M $\frac{1}{2} \mathbf{W}$ high stability.
> R4-600k, $\frac{1}{2}$ W high stability.
> R5-200k $\frac{1}{2}$ W high stability.
> R6-100k $\frac{1}{2} \mathrm{~W}$ high stability.
> R7-60k $\frac{1}{2} \mathbf{W}$ high stability.
> R8-20k $\frac{1}{2} \mathrm{~W}$ high stability.
> R9-10k $\frac{1}{2} \mathbf{W}$ high stability.
> R10-10k $\frac{1}{2} \mathrm{~W}$ high stability.
> R11-470k $\frac{1}{2} W$.
> R12-47k 1 W .
> R13-47k 1 W .
> R14-1k $\frac{1}{4} \mathrm{~W}$.
> R15-10k $\frac{1}{2}$ W.
> R16-2 $\mathbf{2}$ to 4M high stability.
> R2-R10 are 1 per cent tolerance; R11-R16 20 per cent tolerance.
> VR1-200k preset potentiometer.
> VR2-25k (or 50 k ) wotentiometer.
> $\mathrm{C} 1, \mathrm{C} 2-0.05 \mu \mathrm{~F}, 500 \mathrm{tW}$.
> M- $0-500$ microammeter.
> SI-Single-pole 9 -way ceramic single wafer switch.
> S2-Double-pole double-throw toggle switch. Via and V1b-12AU7 or similar (see text).

requirements of the constructor, but an example of a simplified instrument is given in Fig. 3. In this circuit the ranges provided are 0 to 100 V (switch position 1), 0 to. 10 V (position 2) and 0 to IV (position 3).


Fig. 4.-An A.C. test-probe.

## A.C. Probes

A separate probe must be used for all A.C. measurements and it is desirable that different probes should
be used for audio and radio frequency measurements. The circuit of an A.C. probe is shown in Fig. 4. Cl removes all D.C. from the input and the series resistor RI prevents the diode and the capacitance of the coaxial cable from loading the circuit under test to any appreciable extent. The diode D1 shunts all negative-going pulses to the screening, whilst positive pulses are fed down the coaxial cable to the main instrument. The valve grid receives a steady positive voltage which is measured.


Fig. 5.-A suitahle power supply unit. "D" should be $150 \mathrm{~V}, 5 \mathrm{~mA}$, or more.

In the case of the audio probe, the value of Cl may be $0.05 \mu \mathrm{~F}$ or $0.1 \mu \mathrm{~F}$, but a 500 pF , ceramic condenser is much more suitable for the radio frequency probe. If one wished to make high audio and low radio frequency measurements only, it would be possible to do this with one probe by using the compromise value of about $0 \cdot 002 \mu \mathrm{~F}$ for Cl in Fig. 4.

The use of the resistor R1, shown in Fig. 4, not only enables the range to be accurately adjusted, but also prevents the circuit under test from being appreciably loaded by the capacitance of the coaxial cable. The average constructor will not wish to make actual measurements of radio frequencies in volts, -but will only require to compare voltages (e.g. uhen aligning a receiver). It will therefore probably be unnecessary to calibrate the R.F. probe against $a^{\text {a }}$ standard instrument. The audio probe can be calibrated by using suitable $50 \mathrm{c} / \mathrm{s}$ voltages from transformers. and a standard A.C. meter. R1 of Fig, 4 may be adjusted so that the instrument reads either peak A.C. voltages or root mean square voltages, as desired.

An OA81 is a suitable diode for both R.F. and audio probes, but care should be taken that no attempt is made to measure an A.C. voltage which has a peak value greater than about 50 V . The

## TABLE I

VOLTAGE RANGES OF THE CIRCUIT SHOWN IN FIG. I

## Switch

 Position 12

Range (V)
0-1000
0-500
0-250
0-100
0-50
0-25
0-10
0-5
0-1
maximum frequency at which the R.F. probe can be operated satisfactorily is not known, but it should be quite satisfactory for comparative measurements to over $10 \mathrm{Mc} / \mathrm{s}$. The minimum frequency of accurate operation of the audio probe is well below the lowest audible frequency.

## Power Supply

The instrument requires 6.3 V at 0.3 A for the valve heater and approximately 3 mA at about 150 V as H.T. This can usually be obtained most conveniently from a receiver power supply using a suitable series dropping resistor. If the receiver H.T. voltage is 250 V , a $33 \mathrm{k}, \frac{1}{2} \mathrm{~W}$ resistor should be used in the H.T. lead. Alternatively the power supply shown in Fig. 5 could be used.

## Construction

The controls on the front panel are S1, S2, and VR2. The meter itself is also fixed there. Resistors R2 to R10 should be mounted around the switch wafer, S1. The potentiometer VR1 should be mounted on a piece of ebonite or perspex in order to minimise leakage. The insulating material used can be mounted on a hole cut in the chassis.

## The P.W. Signal Generator

(Continued from page 994)
little difficulty should he encountered if the transformer maker's connection details are followed carefully in conjunction with the circuit diagram.

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# A Capacity Tester for Small Values 

THE UNKNOWN CONDENSER IS USED TO ALTER THE TUNING OF AN R.F. OSCILLATOR

S<br>By B. J. Roome<br>OME types of small capacitors are not marked with their capacitance value at all, whilst the values marked on other types--especially the markings on the small waxed mica types-often become obliterated with age. In addition the meaning of the colour coding of some old types of capacitor is not always clear and the colours may not be decisive. If such capacitors are to be used in any apparatus, it is first necessary to ascertain their value. Even if the value is clearly marked, a quick check is worthwhile if the component is not new and will also disclose any open circuited or short circuited component. It is much wiser to test a doubtful component before use than to spend a very long time finding out why a completed piece of apparatus will not function.

## Principles of Circuit

Fig. 1 shows a circuit which can be used for measuring the value of small condensers. V1(a) is an oscillator, working at a fixed frequency of a few megacycles. The exact frequency is unimportant providing that it is constant. The tuned circuit L/C2 determines the frequency of oscillation. The output of the oscillator is fed, via C7, into V1(b) which is a cathode follower buffer stage. This stage prevents pulling of the oscillator frequency by the test circuit. L4 couples the output from V1(b) into the test circuit
which absorbs maximum power from L3 when the tuned circuit formed by L4, L5, C8, C9 and any capacitors under test ( Cx or Cy for future reference), resonates at exactly the same frequency as the frequency of oscillation of $\mathrm{VI}(\mathrm{a})$.
The voltage appearing across the test circuit is detected by the germanium diode D1 and the resulting D.C. current deflects the $0-500$ microammeter M1. C10 by-passes the R.F. so that it does not pass through the meter, whilst VR1 enables the current passing through the meter to be adjusted to a suitable value.
The 12AU7 is perhaps the most convenient miniature valve, but the octal based 6SN7 is quite suitable. Alternatively two separate 6C4's or 6J5's could be used. Almost any small triodes or triode connected pentodes are perfectly satisfactory. The part of the circuit on the right-hand side of the screen in Fig. 1 is placed on or near the front panel with the valve behind it.

## Preliminary Adjustments

A 0-500 microammeter should be inserted between the lower end of R2 and the chassis at the point marked with a cross in Fig. 1. If no reading is obtained at first, reverse the connections of L2. The reading of the microammeter is a measure of the amplitude of oscillation of V1(a). The number of turns on L2 and the spacing of L2 from L1 should be adjusted until the ammeter in the R2 circuit reads about $100 \mu \mathrm{~A}$. The amount of feedback in the oscillator circuit is controlled by the number of turns on L2 and by its proximity to L1. Generally L2 will consist of about seven turns spaced to to $\frac{1}{2}$ in. from L1. After adjustment, L2 should be fixed in position with polystyrene cement. R1 drops the H.T. voltage to about 100 V and also decouples the anode circuit of V1(a).

C 8 and C 9 should then be adjusted (with no connection across the Cx or Cy terminals) until a reading is obtainied on the meter M1. The trimmer C5 should be adjusted for maximum reading on M1 and it should then be set in position with a suitable cement. No further adjustment of the circuit should be required for some years.

## Calibration

A fairly large type of tuning dial (calibrated from 0 to $180^{\circ}$ ) is fitted on the front panel and connected to C 9 . It has a vernier attachment for accurate reading to $0 \cdot 1^{\circ}$, but this is not essential. It is advisable to arrange for the dial reading to be zero when the vanes are fully meshed.
A number of 1 or 2 per cent tolerance capacitors should be used for calibration of the instrument, but 5per cent capacitors will suffice for
approximate work, those used for calibrating the Cx position should all have values between 0 and 150 pF .
C9 should be set so that the dial reading is zero when the vanes are fully meshed. C8 is then adjusted for resonance as shown by a maximum reading of M1. If the meter goes off the scale, the resistance of VR1 should be increased, but when C8 has been set, VR1 should be adjusted so that the meter does not read less than half scale (for accuracy). Place one of the calibrating condensers across the terminals marked Cx in Fig. 1; the meter reading falls as the circuit is now off resonance. Alter the value of C 9 until resonance is again found, care being taken that C8 is not touched. The reduction in the value of C 9 is then equal to the added calibrating capacitance, Cx. Note the reading of the dial connected to C9. Repeat this for other values of calibrating condensers.
A graph of Cx against the dial readings should then be plotted as accurate as possible. The graph should be almost exactly a straight line if C 9 has semicircular plates. The graph should be kept for use with the instrument.

## Operation

Values of capacitors from 0 to 150 pF can be measured in the following manner. Adjust C8 to resonance with the dial reading of C 9 at zero. Connect the unknown, ( $x$ across the terminals and, leaving C 8 unaltered, find the C 9 dial reading at the new resonancc. Using this dial reading and the graph, the unknown capacitance value can be found. The result should be accurate to $\pm \mathrm{lpF}$ or better.

## Larger Values

The maximum value of the measuring condenser, C9, is 150 pF , and therefore capacitance values greater than this cannot be measured by connecting the unknown directly across the Cx terminals. If however, the unknown condenser is placed across the terminals marked Cy in Fig. 1, it will be placed in series with a 150 pF capacitor. The series combination will have a value of less than 150 pF -in fact it will be between 75 pF and 150 pF -and the value of this combination can therefore be measured as before.

Thus it is possible to calculate the value of the unknown Cy , from the following equation

$$
\mathrm{Cy}=\mathrm{Cr} \cdot \mathrm{Cs} / \mathrm{Cr}-\mathrm{Cs}
$$

where Cr is the capacitance of Cy and Cs in series, it is well worthwhile calibrating the instrument for values of Cy greater than 150 pF . This should be done by placing various known calibrating capacitors across the Cy terminals and plotting the values of Cy against the dial readings as before. In this case, however, the graph will not be a straight line. The graph can then be used to obtain values of unknown capacitors directly from the dial readings.

Whilst there is no definite maximum limit to the value of capacitance which can be measured, the accuracy of the measurement will fall as the value of the unknown increases above 150 pF . The practical upper limit is in the region of 2,000 to $10,000 \mathrm{pF}$, depending on the accuracy required and on the accuracy with which the dial can be read.

## Final Check

The accuracy of the instrument and of the graph can be quickly checked by determining the capacity of two condensers separately and then determining their capacity in parallel. The latter value should be equal to the sum of the two separate capacities.


Fig. 2.-A suitable power supply circuit for the tester.

## Power Supplies

The instrument requires a 6.3 V heater supply.at about 0.3 A and an H.T. supply of 200 to 240 V at about 12 mA . This can usually be obtained from a receiver power pack, but if it should be desired to make a separate supply for the unit, the circuit shown in Fig. 2 is quite adequate. The metal rectifier should have a rating of not less than 300 V at 15 mA .

Alternatively it would be possible to use a small battery double triode (e.g. type DCC90 or 3A5) and make the unit entirely portable. The H.T. voltage could be obtained from a 90 V H.T. battery, in which case R1 should be reduced to 2 k . The batteries should have a very long life-almost as great as their shelf life.

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# A D.C. TEST METER 

## A SIMPLE INSTRUMENT FOR MEASURING VOLTAGE, CURRENT AND RESISTAN ${ }^{-}$ WITH EXTERNAL SHUNTS

J.

By B. E. Wilkinson

HIS meter is not intended to be the ultimate in test meters, but rather more as a beginner's instrument or as a second one for quick checks. It has the advantages that it is very easy to make, is inexpensive and the number of ranges available is almost unlimited. This is because the series resistors or shunts used to provide the voltage or current ranges are not installed internally in the meter, and switched in as required by a range selector switch, but are made up as jacks, and are plugged in externally. It is thus possible to provide additional ranges when required. To enable those interested to select their own choice of ranges and provide the correct shunts, etc., the calculations will be given for the general case, and a set of recommended ranges.

## An Ex-Government Meter

The meter, which is intended to measure D.C. current, voltage and resistance to about 5 k , is constructed from a robust ex-government meter, which is intended for checking cells, and measuring resistances approximately. The only indication of type is "Instrument Testing No. 1", which appears at the top, in front of the three terminals and toggle switch. These meters are currently available. The meter is enclosed in a thick, black plastic case, the movement
 itself be ing protected by a circle of thick glass and a metal plate screwed over the front. The scale is graduated in volts to 1.5 or 3.0, milliamps to 60, and ohms to infinity, 5 k being the highest indicated value.

(b)

(c)

The, Meter Movement Before deciding upon the ranges required, it is necessary to know something of the meter movement. On the face for instance it states that it has a resistance of $250 \Omega$.

Fig. 1 (left):
(a) The circuit of the instrument.
(b) Equivalent circuit for the low voltage range.
(c) Equivalent circuit on the 60 mA range.


The completed meter.
This is essential to calculations, but is not sufficient, for the current required for full scale deflections of the meter needle must be known. However, it will be found that a circuit diagram of the instrument as it stands is provided on the top, and is reproduced in Fig. 1a. Now, from the back of the instrument, where operating instructions are given, it is seen that with the toggle switch to (L) voltages to 3 V can be measured across ( - ) and (B). Since voltages are being measured it is clear that the $250 \Omega$ resistor must be in series with the meter and the $27 \cdot 77 \Omega$ shunt resistor must not be connected. Position (L) must therefore be with the switch open. For this voltage range then, the circuit will be similar to Fig. 16, where the second $250 \Omega$ resistor represents the meter movement. The total resistance of the circuit is $250+250=500 \Omega$ and the applied voltage for full scale deflection is 3 , so that by Ohm's law the current flowing will be.

$$
=\frac{E}{R}=\frac{3}{500}=6 \mathrm{~mA}
$$

This then is the current required fully to deflect the meter needle, and is a basic value in the calculations. This value may be checked by considering the currentmeasuring range of the instrument. It will be seen that with the switch in position (R) 60 mA will cause a full scale deflection between ( - ) and (A). This is clearly with the shunt connected but without the series $250 \Omega$. The equivalent circuit is shown in Fig. 1c. Here 60 mA are being measured, but simee only 6 mA may pass through the meter, $60-6=54 \mathrm{~mA}$ must pass through the shunt. The potential difference across the meter will be $\mathrm{E}=\mathrm{I} . \mathrm{R}=6 \times 250=1500 \mathrm{mV}$. This must be the same as the P.D. across R (since they are in parallel), so that the value of $R$, will be, by Ohm's law, $R=\frac{E}{\mathrm{I}}=\frac{1500}{54}=27.77 \Omega$, which is in fact the value of the shunt used. We are now in a position to discuss the required current and voltage ranges and the shunts and resistors necessary to provide them.

Fig. 2 (right):
( $a, b$ and c) How the shunts are calculated.

## Current Ranges

As far as current is concerned, the meter scale is graduated in milliamps to 60 . The ranges considered acceptable for this sort of meter are 6, 60 and 600 mA . By using multiples of 6 , confusion over readings is avoided. Now, since the meter itself has a 6 mA movement, the first range is achieved without a shunt. In Fig. 2a the meter is in parallel with a shunt-which is to be calculated for any current flowing (i). Since the meter must only pass 6 mA , the shunt must pass (i-6)mA. The potential difference is again 1500 mV , and by Ohm's Law R1 will be

$$
R 1=\frac{E}{I}=\frac{1500 \Omega}{(i-6)}
$$

(where $I$ is in milliamps). By substituting for I in this relation it is possible to provide R1 for any

Fig. 3 (right)-Circuit of the instrument when used as an ohmmeter.
desired current range (providing of course it is not less than the sensitivity of the meter movement itself). Thus for a 60 mA range

$$
\mathrm{R} 1=\frac{1500=27 \cdot 7 \Omega}{60-6}
$$

Fig. 4 (right).-Complete circuit arrangement.

And for a 600 mA range $R 1=1500=1500=2 \cdot 252 \Omega$ 600-6 594

## Voltage Ranges

Two voltage ranges are provided on the meter scale, 1.5 and 3 . It is suggested that
 multiples of 3 be considered, i.e.: $3 \mathrm{~V}, 30 \mathrm{~V}$ and 300 V , though the reader is not bound to follow these ranges. To calculate the series resistors required, Fig. 2 b must be consulted,
where the meter is shown in series with a resistor R1 and a voltage to be measured (V). The total circuit resistance (neglecting the source) is ( $250+$ R1), and since for full scale deflection, the current flowing must be 6 mA the potential difference (V), must be given by

$$
V=\frac{6(250+R 1)}{1000}
$$

To obtain R1

$$
\begin{aligned}
1000 \mathrm{~V} & =1500+6 \mathrm{R} 1 \\
\mathrm{R} 1 & =\frac{1000 \mathrm{~V}-1500}{6} \\
R 1 & =\frac{100(10 \mathrm{~V}-15)}{6}
\end{aligned}
$$

(The 1000 factor is necessary to convert milliamps into amps.)
For 3 V range, the relation becomes:

$$
R 1=\frac{100(30-15)}{6}=\frac{100(15)}{6}=250 \Omega
$$

For a 30 V range,

$$
\mathrm{RI}=\frac{100(300-15)}{6}=\frac{100(285)}{6}=4750 \Omega
$$

and for a 300 V range,

$$
\mathrm{RI}=\frac{100(3000-15)}{6}=\frac{100(2985)}{6}=49750 \Omega
$$

For convenience the required values are tabulated in Fig. 2c.
No multimeter should be without a continuity test and ohm range (in fact it is frequently felt that this range is the most useful of all) and since the face is provided with a scale in ohms the opportunity can hardly be missed.
The ohmmeter, which, as will be seen later is built into the instrument, consists of a series circuitmeter, battery, and two resistors, one for adjustment and the other being measured.
Now, on the back of the instrument, instructions will be seen, a note stating that to use the ohmmeter, a cell reading 1.5 V should be connected in series with the resistor under test and the ( - ) and (A) terminals. No adjuster is provided, because the cell is known to be 1.5 V before the measurement is made, and the scale is calibrated for use with this cell and a meter resistance of $250 \Omega$.

Therefore the meter is used as it is for measuring resistance by merely providing a 1.5 V dry battery. Since at first this may provide a little more than 1.5 V , it is well to use a small variable resistance ( $10 \Omega$ ) in the circuit to provide a zeroing adjustment. In Fig. 3, the ohmmeter circuit is shown, if the cell is giving exactly $1 \cdot 5 \mathrm{~V}$, then R1 will be adjusted to zero resistance. Suppose that for a resistance R2 (as yet unknown), a current flow of, say, 5 mA . The total circuit resistance is $(250+\mathrm{R} 2)$. Thus we may write:

$$
\begin{aligned}
\mathrm{E}=\mathrm{I} \times \mathrm{R} \quad 1 \cdot 5 & =5 \frac{(250+\mathrm{R} 2)}{1000} \\
1500 & =1250+5 \mathrm{R} 2 \\
5 R 2 & =250 \\
\mathrm{R} 2 & =50 \Omega
\end{aligned}
$$

It will be seen from the meter scale that a current of 5 mA ( 50 indicated because of the internal shunt), corresponds to a resistance of $50 \Omega$. Since the ranges of the proposed multimeter have been discussed, and the values of resistors required have been established, the actual circuit may now be considered.

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## The Meter Circuit

The circuit is shown in Fig. 4. Here it will be seen that while there is no range selection switch, two jack sockets J1 and J2 are provided. JI accepts shunts across the meter movement, so that current is measured between terminals (-) and (B). J2 accepts resistors for the voltage ranges, the terminals used being (-) and (A). When neither J1 or J2 is being used, the switch may be closed, and resistance measured between (-) and (A). The variable resistor shown is 105 ! and is for zeroing.

## Constructional Details

It is now necessary to modify the instrument, and rewire it using the circuit shown in Fig. 4. The unit is held together by eight 4B.A. screws, four at the front, and four holding the back plate in position. If all of these screws are removed the wiring inside may be disconnected and removed, and the meter movement pressed out, forward. Three spring clips, attached to the meter flange by three 6B.A. screws should be removed and discarded. A thick piece of glass under the front plate and two soft-rubber rings are no longer required and should also be discarded.


A rear view of the complete instrument.
The two resistors wound on a plastic former must be removed, but should be set aside as the resistance wire with which they are wound is required. Measurement of the centre hole in the brass plate which was screwed to the front of the unit shows it to be 2.5 in . in diameter. The meter itself is 2.7 in . in diameter, and since it is intended to fit the meter to the plate, the hole must be enlarged by $0 \cdot 1$ in. all the way round. This can be done quite quickly with a sharp halfround file. When the meter will fit easily through the hole, it should be lined up so that a line drawn through the centre of the scale and the needle adjustment screw is parallel with two opposite sides of the plate. Clearance holes (6B.A.) may now be drilled through the plate to mate with the coriesponding holes in the meter flange. The meter may then be bolted to the plate and the plate fitted back on the case.

Looking into the back of the unit, it will be seen that this modification has moved the meter forward, and thus provided sufficient space for further components. The two jack sockets J1 and J2 may now be fitted. They are fitted, one each side of the case, by means of $\frac{1}{6}$. holes drilled as shown in Fig. 56 which also shows the meter muunted to the front plate.

The 1082 variable resistor for adjustment of the ohmmeter range, is fitted through a $\frac{3}{8} \mathrm{in}$. hole drilled in the top of the case, at the intersection of two lines drawn between the centres of the $(-)$ and $(B)$ terminals and the switch and the (A) terminal. The shank of this resistor should be cut down to about $\frac{1}{8}$ in. and a saw cut made across the diameter to accommodate a screwdriver. The ohmmeter range needs a 1.5 V cell to operate the meter, and this takes the form of a single pen cell held to the bottom of the inside of the case by means of a small Terry clip. This can be seen in the photograph of the back of the instrument.

## Additional Wiring

Wiring should be carried out with single-strand plastic covered wire similar to that removed from the meter initially. Fig. 5 a shows the wiring system. A length of wire is connected between the ( - ) terminal, and the negative point of the meter. Then a wire is jointed to the positive side of the meter and one side of the toggle switch. The other side of the toggle switch is taken to one side of the $10 \Omega$ resistor. The two meter terminals are now connected to the two tags of J1, and the outermost tag of J1 is taken to the corresponding tag of J 2 . Terminal ( B ), is now taken to the positive side of the meter while terminal (A) is taken to the inner tag of J 2 and the negative side of the 1.5 V cell. Finally the centre tag of the resistor is taken to the positive side of the cell. In wiring, unless leads are taken to terminals, they should be soldered. The instrument is now virtually complete and the ohm range can be tested. Set the toggle switch to R and connect a piece of wire from ( - ) to (A). The meter should indicate zero resistance. If it tries to indicate less than zero; i.e. beyond full scale deflection, it can be set to zero by means of the $10 \Omega$ resistor. If it indicates a finite resistance, the cell is down and should be replaced. Tests can now be made using resistors up to about 5 k . The rear plate of the instrument may now be screwed in position. However, it will be observed that engraved on the back are instructions which are no longer relevant, so that the back plate may be put on reversed, the four holes being countersunk slightly to receive the 4B.A. screws.

## The Jacks and Test Prods

To complete the instrument, it is necessary only to provide the shunts and resistors in jacks, and the test prods. Referring to the table of Fig. 2c, we see that for the recommended ranges we require five resistors. For the current ranges the resistors are low, and are made-up from resistance wire. The resistor removed from the meter initially consists of a $27 \cdot 7 \Omega$ and a 250 S resistor. If the $27 \cdot 77 \Omega$ resistor is carefully unwound avoiding kinks, as this piece of wire is required, and measured, it will be found to be almost exactly 8 ft long. Thus:

$$
\begin{gathered}
27 \cdot 77 \Omega=8 \mathrm{ft} \text { or } 96 \mathrm{in} \\
\therefore 1 \Omega=96 / 27 \cdot 77 \mathrm{in} \\
\therefore 2 \cdot 525 \Omega=\frac{96 \times 2 \cdot 525=8 \cdot 726 \mathrm{in}}{27.77}
\end{gathered}
$$

Thus a piece of resistance wire 8.726 in . long, taken from the $250 \Omega$ winding, will provide a shunt for the 600 mA current range. When using this wire, the cotton covering should be removed at each end, and the wire tinned for solderingin. It should be remembered that the actual resistance between the two tags joined by a length of resistance wire does not include that wrapped around the tags, or tinned. Thus the length of wire cut should be slightly longer than that actually required. The jacks to which the resistances are connected are of the standard Post Office type. The resistances are connected across the two connections in the handle of the jack, and the handle replaced. No trouble should be experienced with the series resistors, but the shunts, being made of wire need some support. They may be wound upon a small Tufnol former, with small lengths of tinnned copper wire at either end to provide lead-out tags. The range provided by each jack should be typed on a small label or piece of paper and stuck to the appropriate jack by means of Sellotape. Pieces of paper with "Current" and "Voltage", should be similarly fixed to JI and J 2 respectively, on the meter case, so that confusion does not arise.

Test prods are an essential part of every test meter, and should be made from flexible, stranded insulated wire. The ends connecting to the meter should be soldered to short lengths (lin. or so) of brass wire, the joint being neatly taped. The brass wire will then fit easily in the angled slots cut in the terminals. At


Fig. Sa (left).-Rear wiring details.
Fig. $5 b$ (right).-Side view of the completed test-meter.
the other ends of the leads either brass prods or crocodile clips may be fixed. The brass prods should be sleeved almost to the end, so that one does not hold the bare metal when making measurements.
The instrument should require no calibration, but it is advisable to check it either against a test meter known to be correct, or resistors and cells known also to be correct. To check, for instance, the $3 V$ range it is necessary only to use 1.5 and 3 V cells, while the $30 \vee$ range may be checked with grid bias batteries, car batteries, etc. Any adjustments necessary may be effected by varying slightly the values of resistance in the appropriate jacks.

## QUARTZ CRYSTAL ETCIIING <br> (Continued from page 979)

rate of about $1 \mathrm{kc} / \mathrm{s}$ per minute of immersion has been found with crystals of a frequency of about $8 \mathrm{Mc} / \mathrm{s}$, but a longer etching time per kilocycle frequency rise would probably be required for crystals of lower frequencies.
The crystal should be completely immersed in the etching solution for a comparatively short time-perhaps ten minutes-or' less if a small frequency change is desired. The crystal should be removed from the etching solution with the coated tweezers. washed in two successive jars of water (not hard water) and finally dried in a current of warm (not hot) air. Alternatively, the final wash can be made with pure alcohol which is then allowed to evaporate.
The crystal is then remounted in its holder. If it is a type which is held at the corners only by the metal electrodes, care must be taken that the electrodes are replaced so that the projecting corners face the quartz crystal. The frequency of the etched crystal should be determined as accurately as possible. The rate of etching can be determined from the frequency change and an estimate can be made of the time for which the
crystal should be left in the same solution in order to achieve the desired increase of frequency.
The crystal should be etched for not more than three-quarters of the calculated time and the frequency checked again. It should now be possible to calculate fairly accurately the etching time required to increase the frequency to the desired value. If the final frequency must be extrenely close to the desired value, it is advisable to etch for somewhat less than the calculated time and to check the frequency several times before completing the process. With care the error in the final frequency will not be appreciably greater than that of the measuring device used-certainly less than $1 \mathrm{kc} / \mathrm{s}$.
As etching proceeds, the anode current of the oscillator (Fig. 1) will probably increase somewhat with each frequency check owing to a reduction in the activity of the crystal. The anode current should be noted at each frequency check; an increase of more than a few milliamperes may indicate that the crvstal is useless. Do not forget. however, that a change in the oscillator H.T. voltage may cause a change in the anode current.

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FMT2 is a new tuner in a madern case, green, which can be used for sheld or cabinet mounting, and has space for power supplies if required. Supplied complece with four EF8O valves Hire purchase Deposit $£ 1.14 .6$ and 6 monthly
FMT2 With power. Complete kit
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\&1. 7. 7
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## NEW PRODUCTS AND DEVELOPMENTS

## A precision moulded magnetic tape CASSETTE

NUSIC of just the right kind for hotel lounges, ships lounges. restaurants, fashion houses, etc. or even factories, is provided by the "Reditune" service. Music from a library of over 7.000 titles is provided on magnetic tape which is rented with a playback machine and the required number of speakers giving the effect of a portable orchestra with a repertoire exactly suited to the requirements of the user.

Reditune Lid, is a member of the Rediffusion group of companies.

So that the programme can be changed as needed without any skill on the part of the user, tapes are supplied in ingeniously-designed cassettes which are simply inserted into the machine as easily as posting a letter.

The cassettes are injection moulded by Ekco Plastics Lid. in polystyrene, each cassette comprising four separate mouldings-the body, the spool base, the spoul top and the lid. All except the latter are in high-impact polystyrene, the lid being moulded in transparent polystyrene. The body incorporates two precision-ground brass inserts, a spindle on which the spool turns and a further small spindle to locate an interlocking brake lever. The spool spindle is internally threaded to take the single screw which attaches the lid to the body. The successful operation of the cassette demands freedom from distortion and a high finish with complete absence of thash. A feature which calls for rigidly controlled moulding technique is the need for moulding the spool centre hole to a tolerance of 0.002 in The endless magnetic tape unwinds and rewinds on the spool simultaneously and this involves considerable movement between tape and spool. To facilitate this movement, the base of the spool has 12 radially positioned graphite rods cemented to it, the tape actually resting on these rods during playing. The spool top is moulded separately and subsequently cemented in position. (Ekco Plastics Lid., Press Section, Souhhend- on-Sea, Essex.)

## PERSONAL TRANSISTOR RADIOS

"WO personal transistor radios are now being marketed by Magnavox. Specially-designed printed circuits tor really sensitive long-distance reception, plus hermetically sealed long-life transistors, are included in these new models, retailing at very reasonable prices.

The "Companion" eight transistor radio features a germanium crystal rectifier and a $2 \frac{1}{4} \mathrm{in}$. PM dynamic high fidelity speaker. (Price, 13 gns., including purchase tax.) The "Pocketmate" is a six-transistor version with fine reproduction and performance. (Price, 11 gns., including purchase tax.)

Both are finisined in coloured high impact plastic cases and have leather cases with carrying strap plus private listening earphones. An easel clip allows the Magnavox transistor sets to stand firm on any flat surface. (Magnovox, 129 Mount Street, London, W1.)

## WIRE RECORDER

A N improved and longer playing version of the well-known Minifon P55 pocket-size wire recorder has been introduced by the Office Equip$m \pm n t$ Division $0^{+}$E.M.I. Sales and Service Lid. Known as the Minifon P55L, it now provides a recording/playing back time of five hours. Letters, notes, and memoranda totalling more than 30,000 words can thus be recorded. It is therefore a very convenient instrument for long overseas tours by directors and senior executives, also for busy people -architects, surveyors, civil engineers and plant managers-whose work takes them out and about.

The P55L is the same compact size (4in. by $6 \frac{3}{3} \mathrm{in}$. by $1 \frac{1}{2} \mathrm{in}$.) as the Minifon Attache tape dictation machine, and it slips. easily into an overcoat pocket. The recording medium is very fine ( 0.002 in . gauge) steel wire wound on two small


A new pocket transistor radio.
plastic spools. The fineness of the wire, even compared with $\frac{1}{4}$ in. wide tape, gives the exceptionally long recording time.

The frequency response is $150-3.500 \mathrm{c} / \mathrm{s}$, the input impedance 2 M . and the output impedance 40 k . The low-tension battery is a 1.4 V standard single cell, with a working life of up to 20 hours. The high-tension battery is of 30 V , with a working life of up to 200 haurs. Both batteries are, of course, contained in the compact case. (E.M.I. Sales and Service Lid; Blyth Road, Hayes, Middlesex.)

#  Club News 

## REPORTS OF CURRENT ACTIVITIES

## CLIFTON AMATEUR RADIO SOCIETY

Hon. Sec: C. H. Bullivant, G3D1C, 25 St. Fillans Road, London S.E.6.

Recent meetings have included talks on "DX operating" and "G.P.O. equipment" and a constructional contest which produced 16 entries. On February 24th P. Deacon, G3BCM, will speak on International Radio Regulations.

## MITCHAM AND DISTRICT RADIO SOCIETY

Hon. Sec.: M. Pbaraoh, G3LCH, 1 Madeira Road, Mitcham. The G5UX Key Award was presented to G2BLa who won the $1959 / 60$ contest for this trophy with a score of over 200 points. On January 13th a talk was given on Lifeboat Radio Equipment. The RSGB affiliated Societies Contest was held on February 5th and 6th. The Annual General Meeting will take place on Friday, 24th February.

Future Events:
April 7th-A talk by G2ZU.

## PETERBOROUGH AND DISTRICT RADIO SOCIETY

Hon. Sec.: D. Byrne, G3KPO, Jersey House, Eye. Peterhorough. At their mid-winter rally held in Peterborough Technical College, mernbers had a working demonsiration of new transmitters and receivers by Mr. J. Herries, of Glasgow, who spoke on ssb techniques. Talk-in mobile stations were G3ARS/A on two metres and G3KPO/A on Top Band.

Future Events:
March 3rd-"Aerials". April 7th-Film show. May 5th"Oscilloscopes".

## REIGATE AMATEUR TRANSMITTING SOCIETY

Hon. Sec.: F. D. Thom, G3NKT, 12 Willow Road, Redhill, Surrey. The highlight of the December meeting was a demonstration by BRS 20809 of an ex-R.A.F. morse recorder (made in 1918) which coped quite efficiently with a modern electronic bug-key. The closing date for the constructional contest was January 31st and the entries will be judged at the Annual Dinner on February 1lth. This will be held at Laker's Hotel, Redhill 7 for 7-30 p.m. The tickets can be obtained from any committee member price $15 / 6 \mathrm{~d}$. It is hoped that the chief guest will be Les Knight G5LK who was formerly Secretary of the East Surrey Radio Club for many years.

## SLADE RADIO SOCIETY

Hon. Sec.: C. N. Smart, 110 Woolmore Road, Erdington, Birmingham 23.

The Club Station (G3JBN) at the Church House, High Street. Erdingtom, Birmingham 23, is available for the use of members for constructional purposes. Thursday evening meetings will include informal discussions. operation of the Club transmitter. and instructional Morse classes. A licensed operator of G3JBN will always be in attendance on Thursday evenings. Slow Morse transmissions are radiated on the air each Monday evening from station G3AYJ on $1.9 \mathrm{Mc} / \mathrm{s}$ from $8 \mathrm{p} . \mathrm{m}$. to $8.30 \mathrm{p} . \mathrm{m}$.

Future Events:
February 10 th-A special meeting to consider alterations to rules and a discussion on D/F contests and equipinent for the 1961 season.

February 24th-Sale of surplus equipment.

## SUTTON COLDFIELD RADIO SOCIETY

Hon. Sec.: L. E. R. Hall, 24 Calthorpe Road, Walsall, Staflordshire.

The G3GLQ Trophy was again won by Pat Darrag for a 2 m transmitter. On January 12 th George Collins, among others, gave a talk on the "Applications for Cheap Valves". A discussion on a standardised design for construction as a group project entitled "Build your own Communication Receivers" was the subject of the meeting on January 26th.

Future Events:
February 9th-"Methods of Receiving CW". A discussion of and about CW with demonstration of keys oscillator, B.F.O's etc.

February 23rd--"Crystal Grinding and Etching" by John Symes.
March 9th-Club Station Night.
March 23rd-"Tape Recorders". A talk by Len Hall about some commercial and industrial uses of tape recorders.

## PLYMOUTH RADIO CLUB

Hon. Sec.: R. Hooper, 2 Chestnut Road, Peverell, Plymouth.
Club meetings are held on Tuesday evenings at $7.30 \mathrm{p} . \mathrm{m}$. at Virginia House Settlement, St. Andrews Cross.

Future Events:
February 14ih-Junk Sale.
April 18th-Judging for the Ernie Hillyard Trophy.
May 2nd-Judging for the G5ZT Trophy.

## WANSTEAD AND WOODFORD RADIO SOCIETY

Hon. Sec.: J. Seaman, 67 Beattyville Gardens, Ilford, Essex.
The junior section of the club now meets on Tuesday evenings and not Wednesdays. The senior section meets on Wednesdays. They are then on the air with the club call sign G3BRK. The club shack is now being reorganised with better facilities for constructional work.

## WEST KENT AMATEUR RADIO SOCIETY

Hon. Sec.: H. F. Richarcs, 17 Reynolds Lane, Tunbridge Wells. Meetings are held, unless otherwise stated, at Kent County Council Adult Centre, Culveden House, Culverden Park Road, Tunbridge Wells, Keat. Meetings commence at 7.30 p.m. and end at 10 p.m.

A meeting to plan club stand for Tunbridge Wells Hobbies Exhibition was held on January 13 ih , and a talk on Hi-Fi- Tectniques was given by R. Trevitt on the 27 th .

Future Events:
February 10th-Film Show.
February 23rd-Visit to Crawley Radio Society to hear a lecture on the History of Radio by the Curator of the Scieace Muscum G5CS.

March 10th-Informal meeting,
March 24th-Planning Meeting for National Field Day.

## APPOINTMENTS

Belling and Lee Lid. announce the appointment of B. M. Lee as manager of their industrial group, reporting to the company's general manager, and to the board of executive directors. This is a further step in the divisionalisation of the company into separate industrial, and domestic groups, in a programme of controlled expansion which began with the reorganisation of the sales department earlier in the year. He will be responsible for co-ordinating the production, publicity and sales of the industrial group.

Mr. Lee. who is the son of E. M. Lee, B.Sc., M.I.E.E., director and General Manager, and one of the original founders, has spent a number of years in different departments of the Company, and in other companies in the industry, in training for his new post. His appointment will relieve the directors of a considerable amount of detailed work, allowing them more time for matters of general management and planning for the future.
L. A. Norman, assistant sales manager of Stella Radio and Television Co. Ltd. since May 1960, is appointed sales manager as from January lst, 1961.

During the thirty-two years that Mr. Norman has spent in the radio industry he has become a well-known figure. He joined Philips in their radio sales office in 1928. and after serving in the R.A.F. from 1941 until 1946, re-joined the company in charge of the Century House showroom.

Shortly afterwards he represented Philips radio and television in S.E. and S.W. London and the West End.

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 l2v. 15 amp. with large square cooling tins. $19 / 9$ each.

## EK. GOVT. SMOOTHIVG CHOKES 200 mA., $3-5$ H., 50 ohms, Parmeko 8/9 $100 \mathrm{~mA} . .5 \mathrm{H} ., 100$ ohms $3 / 11$ : 150 mA . 10 H .50 ohms g/8: 80 mA . 20 H ., 900 ohms $5 / 9 ; 120 \mathrm{~mA} .{ }^{12} \mathrm{H} . \mathrm{H}^{2} 100$ ohms 8/9: 50 mA.  <br> EX. GINT. IANS TIIANSITRTHERS Primaries 200-250 v. 50 c.p.s. A.C. 250 v. 80 m.a. 6.3 v. 2 a . <br>  $300-0-300$ v. $60 \mathrm{~m} . \mathrm{a} .6 .3$ v. 2 a. 2 v. a. $12 / 11$ $3000-300$ v. $100 \mathrm{~m} . \mathrm{a} .6 .3$ v. 2 a. 5 ₹. 2 a. $18 / 9$ $350-0-350$ v. $160 \mathrm{~m} . \mathrm{a} .6 .3$ v. 5 a. 5 v. 3 a. $\frac{2}{2} / 9$ 10 v. 10 a. Parmeko Fuliy Shrouded $18 / 9$ 5 v. 10 a. Parmeko <br> 4-4 v. C.T. b-8 amps, Parmekó

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or 56 Morley Street (next to Majestic Bailroom), Bradiord.


Toriginally in 1935 to provide reliable communication even under adverse conditions and soon established a reputation of reliability. Based on a fundamentally sound design and built to precision standards, this high performance receiver was not cheap to produce, ranking among the more expensive representatives of its class.

## Superhet Models

The Senior H.R.O. is no longer manufactured having been superseded by more up-to-date double superheterodyne models of very advanced design.
The author however has had in daily use a Senior H.R.O. for some time in conjunction with a varicty of aerials. This particular model has been available for some time.

## Models Available

The models available on the surplus market are the H.R.O., H.R.O. - M and H.R.O.-MX, all of which apart from slight modifications are fundamentally alike. The valves used in all instances are of the UX based types.

## Valve Line Up

The valve line up is as follows: the first and second R.F. amplifiers are 6 D 6 vaives, the first and second 1.F. amplitiers are 6 D6 valves and the mixer valve is a 6C6. There is a separate local oscillator 6 C 6 valve. The detector AVC diode and first audio amplifier is a 607. The beat frequency oscillator is a 6C6 valve and the power output valve is a 42 .

## Tuning Range

These receivers cover a very wide band of frequencies by means of a separate four-coil unit. The total range being from 10 m to 6000 m . nine coil units being provided. These are of the general coverage type which are not fitted with the extra band spread facilities common to the pre-war standard models.

## Standard Model Bandspread

In the standaril models, bandspread was accomplished by the transfer of set screws from one set of contact plates to another. By this means small padding condensers were introduced in series with the tuning condenser to reduce the turing range. and a parallel trimmer to adjust the range to the desired part of the tuning dial.
This system enabled the user to cover the amateur bands with the more limited spread of the general coverage setting with one coil and with extra band-
spread covering 400 divisions of the 500 division luning dial. The overlap between the two coils being arranged with this idea in view.
That the surplus models are supplied only with nine general coverage coils is not the disadvantage it may appear, as the normal dial spread is sufficient to meet the requirements of the average listener, as later details relative to the main tuning arrangements will show.

## Loter Models

In addition to the earlier models mentioned, all of which use glass valves, later models in which the international types are used are also available. The valve line up being $6 \mathrm{~K} 7 \mathrm{M}, 6 \mathrm{~K} 7 \mathrm{M}, 6 \mathrm{~J} 7 \mathrm{M}$, 6 J 7 M , $6 \mathrm{~K} 7 \mathrm{M}, 6 \mathrm{~K} 7 \mathrm{M}, 6 \mathrm{SQ} 7 \mathrm{M}, 6 \mathrm{~J} 7 \mathrm{M}$, and $6 \mathrm{~V} 6 \mathrm{GT} / \mathrm{G}$, and the models are the H.R.O. ST and 5R. The former being the table model and the latter the rack model.

All were manufactured during the war and are fitted with a $0-1$ milliammeter instead of a calibrated $S$ meter. According to the National Company, 0.5 designates S 9 and 1 mA is 40 dB over S 9 .

## Tuning Ranges

The full coverage of the H.R.O. is as follows Coil Sct

| Type |  | L.F. | Bands |
| :---: | :---: | :---: | :---: |
| J - 50 |  | $100 \mathrm{kc} / \mathrm{s}$ | band. |
| $\mathrm{H}-100$ |  | $200 \mathrm{kc} / \mathrm{s}$ | ban |
| G - 180 |  | $430 \mathrm{kc} / \mathrm{s}$ |  |
| F-480 |  | $960 \mathrm{kc} / \mathrm{s}$ |  |
| E-900 |  | 2050kc/s | bala |


| Coil Set |  |  |
| :---: | :---: | :---: |
| Type |  | H.F. Bands |
| JA-14.0 |  | $30.0 \mathrm{Mc} / \mathrm{s}$ band |
| JB-7 |  | $14.4 \mathrm{Mc} / \mathrm{s}$ band. |
| JC-3.5 |  | $7.3 \mathrm{Mc} / \mathrm{s}$ band |
| JD-1.7 |  | 4Mc/s band. |

This coverage is a most extensive one and reference to the chat mounted on each coil unit enables the operator to quickly cleek frequency. The calibration of each coil unit being very accurate indeed.

## A Useful Suggestion

One of the bughears met by most users of mains driven short wave receivers is frequency drift after switching on. As the coil units in the H.R.O. are below the chassis and well insulated from the heat generated in the well ventilated steel cabinet frequency
drift after a twenty minutes warm-up period is negligible.

## Technical Features

One of the highlights of this receiver is the four gang tuning condenser, reduction gear box and tuning dial. The condensers being of the straight line frequency type of 225 pF capacity.

Instead of a common rotor shaft each condenser is completely insulated from the others. This principle of construction has considerable bearing on the stability of the receiver.

Ceramic trimmers are used in all coil units and maintain their setting, resulting in a very superior L/C ratio with constant gain through the various tuning ranges.

The tuning dial with its $20-1$ ratio gear drive which includes a precision hobbed worm is spring loaded. The direct reading dial is divided into fifty divisions with apertures ten divisions apart. Ten turns of the dial are necessary to cover the full travel of the condenser rotors. This totals 500 divisions providing an effective scale length of twelve feet and alows readings to one part in five hundred. The calibration is nearly linear.

## Controls

Fig. 1 shows a detailed sketch of the H.R.O., H.R.O.-M, and H.R.O.-MX type receivers.

## Panel Controls

There are five controls mounted on the front panel and five switches, in addition to the tuning dial.

Using the phasing control in conjunction with the selectivity control enables the receiver selectivity to be varied according to requirements, and interfering signals suppressed. Band width variation is from $2 \cdot 5 \mathrm{kc} / \mathrm{s}$ to $200 \mathrm{c} / \mathrm{s}$ in the sharp position.

As the official handbook supplied with the Senior H.R.O. receivers at present available provides full details as to setting up and operating the receiver controls including the crystal filter, the author does not propose to ofler further comment on the subject.

Fig. 1. shows the beat oscillator control which also has an integral on/oll switeh. The dial plate is also graduated from 0-10. This control clockwise switches 11 .T. voltage to the anode and screen of the beat oscillator valve and between 0 and 10 provides a frequency variation over a range of $10 \mathrm{kc} / \mathrm{s}$.

White the beat oscillator tunes to the I.F. frequency, when set at approximately 9 , the pitch of the note can be varied as desired.

With the facilities which enable the C.W. listener to vary the pitch of the C.W, note, the beat oscillator is an asset to the amateur phone and S.W. broadcast listener in the location of the DX stations carrier wave.
Immediately above the beat frequency oscillator control is the AVC switch. This is switched to the left for $A V C$ in and to the right for $A V C$ out. The AVC: being switched out for C.W. reception, with the beal oscillitlor switched in and pitch control set initially to nine on the dial plate, later to be adjusted to a pleasing note. gain of the second R.F. stage and the two I.F. amplifiers. This control is also used to zero the S neter in conjunction with an internal preset potentioneter.

An A.F. gain control is also fitted. The inclusion of these two controts enables the operator, by the combination of adiustment, to obtain a most satisfactory signal to noise ratio on both strong and relatively weak signals.

## Selectivity Switch

At the top-right-fand side of the panel is the selectivity switch which tunes the secondary of the first l.F. transformer and is used in conjunction with the crystal filter. When the crystal filter is not in use the selectivity control is used as an I.F: trimmer and set to a point which provides maximum volume and sensitivity.

This control in operation is a very effective onc: so much so that in many instances the crystal filter is not required.

## Phasing Control

Below the selectivity control is the phasing control with an integral crystal filter switch.
The crystal phasing control dial plate is graduated from 0-10. When set at zero the crystal filter is out. When at settings between 1 and 10 the crystal filter is in, and the phasing control is used to balance the crystal bridge.


Fig. 1.-The controls of the Senior H.R.O.

## L.F. Gain Control

Above the AVC switch is the L.F. gain control which also has the dial'plate graduated from 0-10. Above this and slighty to the left is the $S$ meter switch and to the right the headiphones jack.

## The S Meter

The $S$ meter should always be switched out when receiving strong S.W. broadeast signals.
As previously mentioned the $S$ meter is not as

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fitted to pre-war commercial models and including dB above S 9 , but is a $0-1$ milliammeter graduated as follows: $0,0.2,0.4,0.6,0.8,1.0$. This form of S meter is not easy to read at a glance and the operator should graph the scale with one graph from 0 to S 9 and another from S 9 to 10 dB over S 9 . Alternatively the two can be combined in one graph.
Apart from its usefulness as a check on variations in signal intensity, an $S$ meter is an adjunct which can be very useful in comparing the signal gain of one aerial and another be they of the same or different types.

This applies especially to fixed relay controlled doublets and rotary beams.

The H.R.O. S meter circuitry is in the form of a three element bridge all of which are fixed. Adjustment of the R.F. gain control brings the controlled valves to a predetermined value of A.C. resistance and the S meter needle to zero amplification will thus conform to the $S$ meter scale.

## Modification

In some instances competent amateurs have modified the H.R.O. by fitting the more modern octal valves or alternatively miniatures. The UX types fitted are, however, in plentiful supply and as it stands the H.R.O. receiver is an excellent performer more especially if used with an efficient aerial.

The sensitivity available far exceeds that which can be used in industrial locations. The judicious use of the R.F. and L.F. volume controls assuring a satisfactory signal to noise ratio on both strong and weak signals.

## Noise Limiters

Most of the modern communication receivers are fitted with some form of noise limiter, but the H.R.O. models outlined in this article are not.

But while such modifying devices which have been designed are effective where there is strong electrical interference with signals of average strengith, what effect the addition of a noise limiter will have on sensitivity and weak signals is worth considering. This being so, a tried and proved design of noise limiter is advised which should have provision for switching out.

## Aerials

Provision is made in the H.R.O. for the use of single wire fed and twin wire fed aerials. If doublets are used the jumper lead from the Eterminal to chassis is removed from the earth terminal end, and the twin feeders inserted in the aerial and earth terminals. This also applies to the noise reducing type doublets and vertical rod systems which are used in conjunction with matching transformers.

Owing to its wide frequency coverage, high sensitivity and selectivity the Senior H.R.O. will meet the requirements of both the amateur bands and short wave broadcast listener.

## Power Supplies

Unlike many other cormunication receivers the H.R.O. has no built in power supply. Type D power units which include a built in 6in. speaker are of British manufacture and are readily available. Also the original National 697 unit which does not include a speaker.

In addition are the 697AB for 105 to $120 \mathrm{~V}, \mathrm{~A} . \mathrm{C}$., and the 6 V vibrator, 686 S .
For those who wish to use a separate loudspeaker the output is 3 W at a nominal impedance of 7 k , the headphone jack being for high impedance headphones.

## A Versatile Tape Recorder

## (Continued from page 991)

place and before anything else is added. the valve heaters are wired with twisted, P.V.C. covered copper wire. After being put in place the wire should be pushed flat on to the chassis. The leads from V4 out of the chassis should be wired with flexible, plastic covered wire.

If the neon bulb to be used has its own lead it should be mounted noss to save difficulty later when wiring is commenced. The switch, volume and tone controls, coaxial sockets and front panel are now added. A screened lead (with screen earthed to the earthed sections on the switch) should be taken along the angle of the chassis from SA to pin 9 of VI. SAI and SA2 are connected to socket A and SC is connected to socket C. (Care should be taken not to overheat the centre pin of sockets $A, B, C$ and $D$ as the insulation may melt.)

The oscillator coil L1 and C19 are now fixed in place; L 1 being attached by its plastic fixing bolt so that the three contacts of the coil are in a convenient position for wiring. Wire the oscillator section completely, except for the earthed ends of $K 21$ and R22 which will be connected later to the earth-bar.


Fig. 12 (left).-An alternative equaliser
Fig. 13 (right).-The tag-board for the equaliser.

## The Oscillator Screen

Bolt the oscillator screen in place. taking care not to short-out any wires. The next step is to fix the large tag-board to the back wall of the chassis by two 6BA bolts, using nuts as spacers. Using sleeving if necessary, wire the tag-board to valve holders 1,2 and 3 . The equaliser tag-board is now mounted is shown in the diagrams and wired in.
(To be continued)

The Editor does not necessarily agree with the opinions expressed by his correspondente

Whilst we are always pleased to assist readers with their technical difficulties, we regret that we are unable to supply diagrams or provide instructions for modifying commercial or surplus equipment. We cannot supply alternative details for receivers described in these pages. WE CANNOT UNDERTAKE TO ANSWER QUERIES OVER THE TELE. PHONE. If a postal reply is required a stamped and addressed envelope must be enclosed with the coupon from page iii of cover.

## PUSH-PULL AMPLIFIER

S
IR.-Referring to the query of $J$ Haskell (January issue). the value of resistors for EL84 valves is about 270!2. 5per cent tolerance. 3 to 5 W . and the cathode hy-pass capacitors $25 \mu \mathrm{~F}, 12 \mathrm{~V}$.W. A suggestion I would like to make is that only one fixed resistor be used, the other being balanced. This ensures that the valves are balanced. If an ammeter is connected in series with the fixed resistor and hy-pass capacitor, and the cathode of the valve. note the reading on the meter. Next connect the meter in series with the cathode of. the other valve and the variable resistor and the by-pass capacitor, now adjust the variable resistor until the reading on the meter is the same as that for the first valve. The valves are now halanced. If the fixed resistor is $270 \Omega$ the variable resistor value should be about $500 \Omega$.-R. Davies (Nottingham).

## high fidelity

IR,-I find it most annoying that the phrase "High Fidelity" is so mis-used.
Now I see small gramophones (with atrocious reproduction, scratch, etc.) described as High Fidelity reproducers; but I had always been taught that high fidelity apparatus was large. with several speakers and far more than portable.

Now. recently I have seen a cardboard disc with a plastic outside for the gronves (which. incidentally had far worse scratch than the average 78 r.p.m. record. for it was a 45 r.p.m. record) with " Hi Fi" Disc in letters one half of an inch high on it.

Would that people would only use "High Fidelity" in its proper sense.-P. H. (Langport, Somerset).

## one valver

SIR,-With reference to Mr. D. Illingworth's letter in the January issue, the internal resistance of his H.T. battery may cause a misleading voltmeter reading.

If his set continues to work while the voltmeter is connected. he can use with advantage a two cell battery for H.T. - C. H. White (Ilfracombe; Devon).

## CRYSTAL SET RECEPTION

SIR,-with reference to Mr. A. Radmore's letter (P.W. January), I should like to inform him that the coil I used in my crystal set was a home wound, medium-wave type and consisted of 90 turns of $22 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. copper wire, on a 2 in . wooden former. The detector was a germanium diode.

He may also be interested to learn that French, German and other Continental stations have been received at good strength. - D. J. Connolly (Edinburgh).

## MODULATION

SIK.-With reference to J. R. Miller's letter (January issue). the snag in his proposed method of nodulation lies in the low plate efficiency of the final amplifier, the maximum efficiency obtainable being about 35 per cent. This is because the peak instantaneous power in a 100 per cent modulated signal is four times the carrier power. Under carricr conditions (i.e. with no modulation), the final amplifier will be producing one quarter of ifs maximum power output. The plate current, and therefore the power input. will be about half the maximum. The plate efficiency will therefore be only half the efficiency at maximum power. i.e. about 35 per cent. In practice the efficiency may be considerably less, since the amplifier must be operated in class AH or B .

In answer to Mr. Miller's second question. to give an ontput of 150 W an input of 200 W to 250 W will be required (depending on the valve(s), used). 100W to 125 W of audio will be needed for modulation. If, however. Mr. Miller means 150 W input. which would give an R.F. output of 100 W to 120W. 75 W of alldio are required. In order to allow for losses, the theoretical power output of the modulator valves should be about 20 per cent greater than the figures given above.

At 750 V on the nlate a parr of 807 s in class AB2 will deliver 120 W into a $6950 \Omega$ load (plate to plate). At 600 V . the output is 80 W into $6400 \Omega$, and may be increased to about 90W by reducing the load impedance. - D. Price (Cape Town, South Africa).
SIR.-Referring to Mr. J. R. Miller's letter in the January issue, I would like to explain why amplitude modulation is only carried out in the final R.R. amplifier. (Apart from the possibility of F.M. being caused hy coupling or poor power supnly regulation, etc.)

In most transmitters the doubling. amplifying and final stages are usually run in class "C" because of the high efficiency possible using this mode of operation ( 70 to 80 per cent). If modulation of intermediate stages is attempted, the

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| tapped 4 7.4 a. Rectifer 8.3 v. 1 а. $Б$ <br> 2 a or 4 v .2 a ditto, $350-0-350$ | 22/6 |
| MLNIATURE, 200 จ. $20 \mathrm{~mA}, 6.3$ *. | 10/6 |
| MIDGET, 220 \%. 45 mA. 6.3 v. 2 | $15 / 6$ |
| SMALL, $2.20 .0-220,50 \mathrm{~mA}, 6.3$ | 176 |
| STD. $250-0.250 .65 \mathrm{~mA}, 6.3$ v. 3.5 | $17 / 8$ |
| HEATER TRANS. 6.3 v. $1 /$ A 7/6, 3A | $10 / 6$ |
| Ditto 1.4, 2, 3. 4, 5, 6.3 จ. 1i A | $8 / 6$ |
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modulated R.F. waveform would be extremely distorted since the output of a class C amplitier is not directly proportional to its input.

Modulation of an early stage, followed by amplitication is possible however if the succeeding stages are operated in linear class B, but the etficiency of such a stage for a modulated R.F. input is only around 35 per cent, i.e. to give the same output in the final amplifier as a corresponding class $C$ stage, the power drawn from the supply would have to be doubled (assumed to be within the valve's ratings).
To modulate 150 W of R.F. the power output from the modulator valves should be around 85 to 90 W to compensate for modulation transformer losses, and this can easily be supplied by 807 's in AB2 with 550 V on the anodes.-M. D. Pollard (Chertsey)

## RADIO MOSCOW

SIR,-I read with interest both Mr. P. J. Kennett and Mr. E. Haines's letter's (January 1961) concerning Radio Moscow. Mr. Haines in his letter states that the Moscow transmitters are the most powerful in the world, this is not so. In the 49 m band Moscow broadcasts on about 20 wavelengths on three of these wavelengths, to be precise $49.67,49.59$ and 49.42 m , the power output is 100 KW . In the other short wavebands, Moscow commands many wavelengths, e.g. in the 25 m band Moscow broadcasts on 23 wavelengths, and the most powerful output on one of these wavelengths is 100 KW on 25.17 m in fact.

Many of the American slations are twice as powerful, e.g. radio stations KRCA, at Dixon, and WLWO at Cincinnati.
In the longwave band, on 1734 m , Moscow's output is $500 \mathrm{KW}, 100 \mathrm{KW}$ more than our own Light programme. on $1,500 \mathrm{~m}$. Even this phenomenal output cannot be compared with the V.O.A. transmitters. at San Fernando, in the Phillipines, and at Okinawa in the Ryukyu chain. These stations broadcast on 263 m . and 254 m respectively, and both their power outputs are $1,000 \mathrm{KW}$ ! This is twice as powerful as the known most powerful Russian transmitter.-T. Roeves (Alcester).

SIR,-With reference to the letter of E. Haines (January 1961), I should like to state that the transmitters employed by Radio Moscow ate not the most powerful in the world. The Voice of America tills this position with three transmitters each having an output of one million watts located at Munich ( $173 \mathrm{kc} / \mathrm{s}$ ), Okinawa ( $1180 \mathrm{kc} / \mathrm{s}$ ) and Manila ( $1140 \mathrm{kc} / \mathrm{s}$ ). It is possible that Radio Moscow's short-wave transmitters are more powerful than those of the Voice of America whose most powerful have an output of 200 KW . However, the Voice of America is currently building a new shortwave station in North Carolina which will house six 500 KW shortwave transmitters which will be in operation by the end of 1962.

Reader P. J. Kennett (January 1961), has already stated the correct facts which need not be duplicated here concerning the relay of Radio Moscow by other stations. However, it is of interest to note the following facts about the Voice of Anerica relay stations:-
There are nine relay stations (excluding those in the U.S.A.) of which eight are land-based. The
ninth is situated off the island of Rhodes in the Mediterranean and takes the form of the United States Coast Guard vessel "Courier". The others are situated in Germany, Great Britain, Grecce, Ceylon, Ryukyu Islands and Philippine islands.

Those Voice of America stations which announce a call sign are located in the U.S.A.- stations WBOU, WDSI, WLWO and WGEO in New Jersey, New York and Ohio and KNBH and KCBR in California.

The only links which the Voice of America studios in Nashington have with the relay bases overseas are via the transmitting stations in the U.S.A. Should trans-Atlantic or trans-Pacific communications be impossible due to prevailing atmospheric conditions the relay bases are obliged to broadcast recorded programmes from tapes.-B. J. Ayres (Chessington, Surrey).

## TRANSISTOR HOLDERS

SIR,-May 1 add a few comments to some of the letters and opinions expressed in the current issue of Practical Wireless.
E. M. Krell asks why nobody has produced a transistor nolder to avoid soldering. If he cares to look through the advertisements in this or most of the other publications of a similar nature he will see then offered time and again.
F. Keating asks how to play a tape backwards. To do this he should turn his tape over as if he was going to play back the other track and then twist it so that the glossy back of the tape is running agains! the playback head instead of the dull-coated side.-J. G. Watt (Yorkshire).

## MODULETION

SIR,-Referring to Mr Miller's letter in the January issue of Practical Wireless, I would like to say that modulation of an oscillator is very undesirable. It is duc to F.M. that always occurs when amplitude modulation is applied directly to the frequency-controlling stage (the oscillator).

In America, at least. amplitude modulation of an oscillator is permitted only on frequencies above $144 \mathrm{Mc} / \mathrm{s}$. -V. Th. Kjartansson (Reykjavik, leeland).

## CRYSTALS

Sir,-In Fehruary Practical Wireless there is a letter from T. L. Mence who requires some Hertzite crystal. I was connected with the wireless trade from 1928 to 1938 and I have a small supply of Hertzite that we used to put into glass tubes for salcs purposes.

If T. L. Mence or any other readers would like a piece (so far as my small slock allows) and cares to contact me I will let them have some if they would send a stamped addressed envelope. I have also been an amateur radio fan on and oft since about 1922 and took the first Practical Wireless. So far back, anyway. that I can remember an issue that gave instructions for running a radio from gas.

I was very interested a month or so agn when Thermion looked back. I also remember the glass tube with practically all of a 3 valve set enclosed nade by a German firm at about 1932.B. ANDREW ( 37 Grimsbury Drive, Banbury, Oxfordshire).

## PLAYING A TAPE BACKWARDS

SIR,-With reference to the letter "Playing a record backwards," February edition, I have two solutions. Either record on a twin track taperecorder then remove the spools, placing them on a four track tape-recorder, one of these tracks will play the recording backwards; or record as normal and then place the left spool on the right and right on left, pick up right spool and turn over (the twist in the tape take comes before the recording head) and play.-M. Brickwood (London, E7).

SIR,-I was interested to read, in the February issue, how Mr. Keating played a gramophone record "backwards," and perhaps I can help him in his efforts to play a tape recording in this manner.

He nust first make a normal recording near the end of a reel of tape. The tapes are then "changed over" so that what was once the "feeder" spool becomes the "take-up" spool. The other spool (which now "feeds"), is forced to move clockwise instead of anticlockwise as it would normally do.

Although this means that the tape can now be played " backwards", it also means that the tape is being played with the insensitive side towards the head and consequently a slight lessening of volume is also noted.-J. B. DE BOER (Huddersfield).

SIR,-I would like to let Mr. Keating know how to play a tape recording in reverse. The only disadvantages are that the output of sound is reduced, and that the higher treble frequencies are lost.

Having placed full reel of tape on the feed spindle as normal, turn it over and replace on spindle. Take the leader tape and join it to the take-up spool in the normal manner. The tape can now be played in reverse, but do not attempt to rewind as the tape will unreel from the spool.

Incidentally, if the reversed tape spools are interchanged to give the same pattern of feed and take-up, then the recorded sound will be heard as normal.

1 hope this will enlighten all interested parties. -D. Hartiand (Bournemouth).

## SHORT-WAVE LISTENERS' LOG-5

It is interesting to hear the news in Special English, radiated each weekday except Saturday by the Voice of America at 19.00 and 20.00 G.M.T. Frequencies and wavelengths used include: $21.5 \mathrm{Mc} / \mathrm{s}, 13.95 \mathrm{~m} ; 15.2 \mathrm{Mc} / \mathrm{s}, 19.73 \mathrm{~m} ; 11.76 \mathrm{Mc} / \mathrm{s}$, $25.5 \mathrm{~m} ; \quad 9.635 \mathrm{Mc} / \mathrm{s}, \quad 31.41 \mathrm{~m} ; \quad 9.62 \mathrm{Mc} / \mathrm{s}, \quad 31.19 \mathrm{~m}$; $9.52 \mathrm{Mc} / \mathrm{s}, 31.51 \mathrm{~m}$ : and $6.185 \mathrm{Mc} / \mathrm{s}$, or 48.5 m . Broadcasts in the international language Esperanto may also be heard, some of the transmissions being as follows: Rio de Janeiro, 30.78 m and 25.48 m , 09.00 to 09.30 G.M.T. Sundays; Sofia, 41.35 m , 39.11 m and $3093 \mathrm{~m}, 19.00$ to 19.25 G.M.T. Sundays; Juiz de Fora, $60.91 \mathrm{~m}, 00.45$ to 01.00 G.M.T. Mondays; Bern, 48.66 m and $31.46 \mathrm{~m}, 17.35$ to 17.40 G.M.T. Mondays: Rome, $50.34 \mathrm{~m}, ~ 41 \cdot 24 \mathrm{~m}$ and 30.9 m .17 .55 to 18.10 G.M.T. Tuesdays.

Listeners who are badly troubled by untunable interference, or lack of good signal strength, may find that a simple dipole solves these problems. With a single-wire aerial. feeding the receiver from one end, static is carried to the set with the signal. But with a dipole, static is in phase in both halves of the aerial, and thus largely cancels out on reaching the twin feeder. The desired signals, on the other hand, are out of phase in the two halves of the aerial, and are thus carried to the receiver. In addition, a dipole does not have the broad-band reception characteristics of a single, end-coupled aerial wire, and this also reduces general interference very considerably.

A dipole can be made from $7 / 22$ or similar aerial wire. Its total length is 0.95 of a half-wave -or about 66 ft for the 40 m band. This total length is cut in half, and an insulator is inserted. From this insulator a twin-lead or coaxial cable is taken to the receiver. One conductor of this feeder thus goes to each 33 ft portion of the aerial. The whole is supported in the usual way with insulators at each end of the two 33 ft horizontal top wires, the feeder descending from the centre to the receiver. Very good results will also be obtained on 20 m and 10 m bands.


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## Practical Wireless

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The index letters which precede the Blueprint Number indicate the periodical in which the description appeared. Thus PW refers to PRACTICAL WIRELESS; AW to Amateur Wireless and WM to Wireless Magazine.
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## SPECIAL NOTE

THE following blueprints include some pre-war designs and are kept in circulation for those constructors who wish to make use of old components which they may have in their spares box. The majority of the components for these receivers are no longer stocked by retailers.

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The PW Monophonic Electronic Organ

## TELEVISION

The PT band III converter ... - $1 / 6$

[^6]


[^0]:    IR5, IS5, IT4, 3S4. 3V4, DAF91, DF91. DK91, DL82, DL94 6K7G. 6K8G, 6Q7G 6V6G

[^1]:    COMPONENTS LIST FOR POWER SUPPLY UNIT.

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    R1-22042 $1 \mathbf{W}$.
    R4-4.7k $\frac{1}{2} \mathbf{W}$.
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    Condensers:
    $C 1-350 \vee W$, any value between 16 and $100 \mu \mathrm{~F}$. Electrolytic.
    C2-As for C1. (These are both obtained in one can.)
    C3, 4, 5 and 6-Each 350VW 8 or $16 \mu \mathrm{~F}$. Electrolytic.
    Chassis: $6 \mathrm{in} . \times 4 \mathrm{in} .2 \frac{1}{2} \mathrm{in}$. high of tin plate or aluminium. (Prototypes use baking tins. The removable bottoms are soldered in place and form the top.)
    Switch: S1 Arcolectric T600.
    Transformers: Tr1. Input to suit mains (or 200/220/240V tapped). Secondary $0-250 \mathrm{~V}$ or $0-300 \mathrm{~V}$ at about 60 mA and $6-6 \cdot 3 \mathrm{~V}$ at about 2 A . Note: Transformers with a centre tap, i.e. $\mathbf{2 5 0 - 0}-\mathbf{2 5 0 V}$ secondary can be used, but one 250 V lead must be cut short or taped and must on no account be used. This type was not used in the prototypes.
    Valve: V1, $6 \times 5$. International Octal Valve base.
    One or two tag strips, 5 tags not earthed, two earthed.
    Warning Lamp: Arcolectric SLB2/1/3 from dealers or Messrs. Arcolectric (Switches) Ltd., Central Avenue, West Molesey, Surrey.
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