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AMEMBER of Parliament, during the recent debate on interference suppression in the House of Commons, gave a warning of the possibility of exploitation of the regulations by spivs. He said that it would be comparatively easy for an unscrupulous person to knock at the door, say that there had been complaints of interference, and then proceed to sell some kind of phoney, and perhaps useless, piece of apparatus.

The reply was that the Regulations would be enforced only when the interference was unreasonable and when the owner would not take action of his own accord. It is probable that only in a small minority of cases would it be necessary to do this.

The G.P.O. analysis of interference complaints shows that the majority of complaints relate to interference with television. During the year ended January 20th, 1955, 55,566 complaints were received regarding interference with sound radio and 85,636 relating to interference with television. It is our view that interference with sound radio by television receivers is more widespread than these statistics would indicate, for the simple reason that most people now own a sound as well as a television receiver and may be reluctant to complain about interference in case one or other of their own receivers is also causing interference.

## THE RADIO SHOW

$T$ Earls Court from August 24th to September 3rd. This and our associated journals will, as hitherto, be represented on our stand. We shall as in past years produce a special design to signalise the occasion. We are at present experimenting with frequency-modulation circuits, and we mention this because of the large number of letters we have received on the subject of F.M. We hope this year that there will be a greater display of components for amateur set builders, and that some of the smaller suppliers of kits and parts will be represented at the Radio Show.

## ABOUT CLUBS

A
S a result of the remarks of a contributor on the subject of clubs, we have received a large amount of correspondence from members of similar organisations seeking our advice. Our
advice is brief and to the point. It is that no one should join a club unless it is run on constitutional lines, where the members themselves have a say in the conduct of the club's affairs and are able to elect their officers, and their club officials each year. In general, proprietor clubs, that is clubs founded by some individual or some group of individuals for the purpose of profit, should be avoided. They. are not clubs properly so described but businesses. It is quite clear from the correspondence that some of these clubs are breaking the law, and that their sponsors are either unaware of or deliberately abrogating the provisions of the various Acts relating to clubs, particularly the Business Names Act.

Our contributor this month postulates a hypothetical case in which someone could start a club, install himself as the proprietor, attract a large membership, and therefore a large annual revenue, and give very little of it back in the form of service to the club. The members would have no redress, except to resign. Where a club claims national or international standing, it should undoubtedly be run on the lines we have already laid down in this journal.

We make these comments because some members of the club in question have stated that they are satisfied with the present order of things. But is the law being obeyed? We repeat what we have stated in other issues, that it is absolutely necessary for a balance-sheet to be published and that such balance-sheets should be independently audited.

This journal has always acted as a watchdog for its readers, and in the 22 years of its history has investigated a large number of clubs, some of whom were found to be specious and merely run for profit. It is because of complaints we receive that investigations are made, and we hope that readers who are members of any club will continue to keep us informed when matters arise which do not comply with club practice.

We do not intend to imply that all clubs are not properly run, but very few are. This may be due in the large majority of cases to incompetence or ignorance, and once the defects are pointed out matters are often remedied. Proprietary clübs are seldom satisfactory and accurate investigation impossible, because the books are not kept in accord with book-keeping practice.-F. J. C.

# Round the Ulorat Wiretess 

## Broadeast Receiving Licences

THE following statement shows the approximate number of broadcast receiving licences issued during the year ended February, 1955. The grand total of sound and television licences was 13,916,246.

| Region | Number |
| :---: | :---: |
| ondon Posta | 1,464,333 |
| Home Counties | 1,419,552 |
| Midland | 1,148,320 |
| North Eastern | 1,523,698 |
| North Western | 1,173,445 |
| South Western |  |
| Wales and Border |  |
| Counties | 78 |
| Total England and Wales 8,271,318 |  |
| Scotland ... | 1,019,013 |
| Northern Ireland | 218,522 |
| Grand Totals |  |
| Grand Totals | 9,508,853 |

## Head of Religious Broadcasting

THE BBC has announced that the Reverend Roy McKay has been appointed Head of Religious Broadcasting as from June 20th.


A technician replaces an output unit into the radio distribution unit of a channel used by controllers sfrcaking to aircraft. (See "Air Traffic Control".)

## By "QUESTOR"

The present Head of Religious Broadcasting is the Reverend Francis House, O.B.E., who was appointed in 1947. His predecessors were the late Dr. F. A. Iremenger (1933-39) and Dr. J. W. Welch (1939-46).

## BBC Concerts

D
URING May the BBC will present a series of seven concerts to be given on. Wednesdays and Sundays by the BBC Symphony Orchestra and Chorus at the Royal Festival Hall. Doctor Bruno Walter will conduct four of the seven concerts, the dates of which are May $11,15,18,22,25$, 29 and June 1.

Air Traffic Control
A NEW Air Traffic Control Centre for Southern England, which has been built at London Airport, recently began operations.

It replaces the Air Traffic Control Centre at Uxbridge, Middlesex, which was separated from the longrange radar unit at London Airport by five miles of telephone wires. The main feature of the new centre is that for the first time the air traffic controllers work alongside the radar unit. This represents a fundamental change. Radar is now accepted as an essential part of the cont rol system, rather than an adjunct, and it is expected that the new integrated centre will handle the ever-increasing volume of air traffic over Southern England with greater speed and efficiency than was possible before.

The heart of the new centic consists of a control room 60 ft . by 50 ft . with a "service space" of the same size immediately below it. Only those items of radio and radar equipment required by the controlling staff are located in the control room itself - the only radar equipments to be seen, for instance, are desk-mounted units of the shape and size of a table-model television set. Associated equipment has been placed immediately below or in nearby equipment rooms and is easily reached for maintenance purposes without disturbing the operating staff.

## G.E.C. Atomic Energy Plans

THE General Electric Co. Ltd., announces that Mr. R. N. Millar, lately chicf mechanical engineer to the British General Electric Co. (Pty.) Lid., Australia, has been appointed to take charge of its newly formed Industrial Atomic Energy Section.

## Cyprus Broadcasting Station

$\mathrm{O}^{\mathrm{N}}$ April 3rd came news that the Marconi 20 kW tansmitter at the Cyprus Broadeasting Station at Nicosia had been damaged by saboteurs and was out of operation.

An emergency plan on an around-the-clock basis was rushed into action at Marconi's Chelmsford works. Forty-eight hours later a complete 2 kW broadcasting transmitter was in packing cases ready for shipment.

The transmitter was flown 10 Cyprus the next day, accompanied by Marconi engineers. Marconi's stated that, barring unforescen contingencies, the equipment should be operational within a short time after arrival.

## Marconi Mast Radiators for Corfu

 IN the face of severe German competition, Marconi's Wireless Telegraph Co., Ltd., has secured an important order from the Greek Broadcasting authorities.The order, which was obtained through the company's Greek agents, Messrs. P. C. Lycourezos, Ltd., is for two 148-metre (approx. 485 ft .) mast radiators of galvanised lattice steel, complete with R.F. transmission line of the five-wire unbalanced type, aerial matching
equipment, an earth system, test apparatus and spares.

- The masts are to be installed on the island of Corfu, at a coastal site about 3 kilometres south-west of the town of Corfu. The aerial matching equipment will be coupled to an existing 50 kW transmitter.

The installation work is to be carried out by Marconi engineers.

## Invisible Hearing Aid

IT is reported from the U.S.A. that a new hearing aid has been developed which is entirely invisible. Made to resemble a pair of horn-rimmed eyeglasses it incorporates over 200 component parts in a standard width and weight spectacle frame, and a thin, colourless, flexible tube lin . in length leads direct to the ear from the bow. The microphone is in the frame directly behind the ear. The miniature battery lasts 180 hours.

## F.M. Litigation

THE long-standing action against R.C.A. has now been settled by the payment of $1,000,000$ dollars by the R.C.A. to the estate of the late Major Armstrong. The action was commenced in 1948 by Armstrong, accusing the R.C.A. and N.B.C. of infringement of five of his basic patents on frequency modulation.

## Hi-Fi Demonstration

A NEW idea in hi-fidelity sound demonstrations, named "Soundorama" took place in Washington. A 90 -piece orchestra played a selection which was tape recorded and then played back through a most ingenious network. For this, ten 50 -watt amplifiers were used, and ten three-way loudspeaker systems mounted on the stage with the orchestra. The audience were thus able to compare the original with the played back recording.

## Inventor of Radar Caught

SIR ROBERT WATSON-WATT, the noted radar pioneer, was 1rapped in Kingston, Ontario, whilst speeding. He was fined $\$ 12.50$, and it was stated that the policc checked his speed by means of a radar installation.

## Ampliphase Transmitter

A NEW system being used by the modulated signal to be obtained
with only a few watts of audio power instead of the 35 kW normally required. The system uses phase modulation principles to produce standard amplitude modulation. Its circuitry permits two phase-modulated amplifiers to produce a combined power equal to the output of much larger AM

Sutton Coldfield Television Transmitting Station in which post he has remained until taking up his present appointment.

## Radio and TV Sales

THE second of the interim statements of retail sales of radio and television receivers derived


One of the radio officers of the new 20,000-ton Shaw Savill passenger liner" Sowthern Cross" at work with the Marconi Marine equipment in the radio room. He is seent tuning the "Mercury" receiver, while other equipment shown "comprises (l. to r.) the "Worldspan"" main transmitter, " Reliance" emergency transmitler. "Electra" receiver "Alert" emmergency receiver and " Allokey" antomatic keying unit.
transmitters, resulting in a saving of operating costs of up to 50 per cent.

## BBC Engineering Division Appoint-

 mentTHE BBC announces the appointment of Mr. T. P. Douglas, M.B.E., as Engineer-in-Charge, Sutton Coldfield Television Transmitting Station.

Mr. Douglas joined the BBC in 1938 as a junior maintenance engineer at the Daventry Transmitting Station.

During the war he held a commission in the Royal Corps of Signals and was, for $3 \frac{1}{2}$ years, a prisoner of war in Siam. He was awarded the M.B.E. in 1948 in recognition of his services while in captivity when he was actively engaged in radio work.
Mi. Douglas returned to the BBC at Daventry in 1946. He was transferred to Kirk o'Shotts Television Transmitting Station in 1951 and in 1953 became the Assistant Engineer-in-Charge of the
from the British Radio Equipment Manulacturers Association monthly market surveys shows that, as compared with January, there was vintually no change in the sales during February of radio receivers and radiograms, but television sales fell in comparison with January by approximately 10,000 .

Estimated National Retail Sales Radios and Television Month Radiograms Receivers January $1955 \quad 122,000 \quad 97,000$ February 1955 122,400 87,200

An approach is being made to over 1,000 additional dealers requesting their co-operation in the extension of the B.R.E.M.A. sample. This extension has been designed to obtain returns for trade in March, and monthly thereafier, from a sufficiently large and representative cross-section of traders to enable reliable regional estimates to be made and for stocks and sales to be analysed, e.g., showing sales for cash and hire purchase or creudit.


A GOOD AMPLIFIER FOR UNDER £5

By A. Selwood

As for valves, practically any good pentode and triode will do ; in the circuit of Fig. 1 an SP61 amplifier and a triode-connected EF50 splits the signal.
The only other feature of interest in the circuit is the use of the 6 volt RK 34 as the double-triode output valve. The RK34 was chosen for no other reason than that it may be bought for just half a crown. Originally it was designed as an R.F. power output valve capable of handling 20 watts, so there is little likelihood of it overloading in the present circuit. Another RK34 with its grids tied to their separate anodes is used as the full-wave rectifier.

The feedback loop and the scratch filter are elementary, and, some may say, rudimentary. Nevertheless they function quite well.

Most "shop" mains transformers will need an extra volt wound on to their rectifier windings, since the RK34 has a 6 volt heater. In most cases this is very easy because the rectifier windings are put on last, at the outside of the coil. But do make sure ! Loose H.T. windings are an unpleasant sight, and 38 s.w.g. is tricky to handle.

## Price List

The price of parts list is complete, and compiled from the catalogue of Messrs. Radio Supply Co. Ltd., Leeds. The items marked with an asterisk

## The Circuit

The heart of this amplifier is the semi-starvătion operated pentode, and the difect coupling to the phasesplitter. The anode of the first valve works about a steady voltage of 100 volts ; this state of affairs being brought about by appropriate choice of the anode load resistor and of the screen potential for V1. Now it so happens that a positive bias of 100 volts or so is required by the phase splitter, which then draws its four or five milliamps quite happily. The simple nature of the coupling and a typical set of operating conditions are shown above. The load resistors of the phase-splitter must be of exactly equal value. They may best be matched, if they are the latge two-watt type, by filing a slot into the resistor of lower value, and deepening it until the resistors are the same value. A touch of hot shellac will keep moisture out of the cut.


Under view of the complete amplifier.
do not appear to be readily obtained elsewhere. When the amplifier is complete, it is essential to check that the anode voltage of V 1 is within 10 volts of 100 volts positive. If it is not, a slight amount of
juggling with the value of R6 will be necessary. Voltages and currents at other parts of the circuit are indicated below; they were measured with a $50 \mu \mathrm{~A}$ meter.



The sound output is more than sufficient for a room 20 ft . square, and the gain is such that the amplifier overloads at maximum volume and 400 mV input. The quality is good to listen to : it cannot be
lished in these pages, whilst we are at all times willing to assist ${ }^{\circ}$ readers who find themselves in difliculty in getting satisfactory results from apparatus built from the descriptions in these pages. we are

## PARTS PRICE LIST

*Mains Trans. $\mathbf{3 5 0 - 0 - 3 5 0}$ v. 70 mA ; 6.3 v. 2 a. : 5 v. 2 a. (needs s'ight modification)
Choke-50 hy 50 mA (or near)... $\quad .$.
*O/P Trans.-push-pull 8 watt 6v. 6 to $3 \Omega$
C8 + C9-16 $/ 1 \mathrm{~F}+16 \mu \mathrm{~F}$ Dubilier electrolytic 500 v .
$\mathrm{C} 3+\mathrm{C} 4-5 / / \mathrm{F}+8 / / \mathrm{F}$ Dubilier electroIytic 500 v .
*V3, V4-RK34 at 2/6
$\begin{array}{llll}\ldots & \ldots & 3 & 9\end{array}$
*V3, V4-RK34 at
V1-SP61 at $2 / 9$
V2-EF50

## HARDWARE

1—chassis 16 s.w.g. $12 \mathrm{in} . \times 8 \mathrm{in} . \times 2$ 2 in .
1-Switch Arrows T.S.P., 250 v. 1 a. ..
1-British screened top cap
1-Fuseholder and 250 mA fuse
*2-UX 7-pin ceramic bases RK34 at 10id.
1-I. O. valveholder
1-B9G ceramic vaiveholder
4-plain top clips
1-tag board
Sundries
s. d.
$18 \quad 9$
9
$7 \quad 11$
59
$\begin{array}{ll}3 & 9 \\ 5 & 0\end{array}$
$2 \quad 9$
9

# The Radio Components Show 

## DETAILS OF SOME OF THE EXHIBITS AT THE R.E.C.M.F. EXHIBITION IN LONDON

TTHIS year's exhibition of component parts showed very little which was new in the radio line, although, as readers of our companion paper will see, several new television items were on show. In the main, however, the exhibition merely consisted of well-known lines plus one or two new versions of popular items. The greatest part of the new developments was concerned with the printed circuit technique, which no doubt in time will revolutionise the construction of all forms of electronic apparatus. T.C.C. had some interesting items in this connection, as they are now specialising in this particular branch. These included:


Front view of the new Plessey 5-channel pre-selector.
Several high quality amplifiers.
Radio receivers of the A.C./D.C. and battery types.
Portable with contained battery.
TV tuners for Band I and Band III.
Aerial filters for TV receivers.
Eross-over networks.
I.F. transformers for TV.
J.F. transformers and amplifiers for TV.

Transistor cemputor panels.
Telephone distribution panels.
Flexible circuits, etc., etc.
Other printed circuit developments were seen on the Belling Lee Stand, and one of these is here illustrated. It is a circuit connector. These are designed for easy installation, and can be mounted side-byside in multiples of four connector strips to produce a practically unlimited number of contacts. They can be "stacked" with spacers between for component mounting. They can be mounted direct to plates, printed or otherwise, and are suitable for mounting on ceramic and laminated plastic printed circuits.

In connection with the development of these printed circuit applications the Multicore Company have developed the special solder which they showed at the last exhibition, and now the complete range at present offered for the soldering of printed circuits is as follows:
P.C.1. Multicore Abrasive Degreaser.
P.C.2. Multicore Dip Cleaner.
P.C.10. Multicore Activated Surface Preservative.
P.C.20. Multicore Special Non-corrosive Liquid Flux.
P.C.21. Oil-free Special Non-corrosive Liquid Flux.
P.C.31. Multicore Alloy for Printed Circuit Solder Baths.
P.C.35. Ersin Multicore 5 -core Solder for hand soldering the printed circuits.
P.C.38. Special Multicore Alloy Service.
P.C.40. Multicore Anti-oxidant Oil.
P.C.50. Multicore Finishing Enamel.

## Trade Items

As the largest manufacturer of components to the trade, the Plessey Company naturally had the widest range of separate items, and as many of these are used by other manufacturers in their products, an inspection of the stand gave a very good all-round idea as to what is available. Apart from the special television apparatus, two new pre-selectors, the PV98 and PV98A, were shown. These are designed for use in conjunction with a single-channel fixed frequency H.F. receiver, such as the Plessey PR53A, PR53C or PR51C, to provide switched selection of any one of five pre-set crystal-controlled channels in the frequency range $2.7-27 \mathrm{Mc} / \mathrm{s}$.

The PV98 is fitted with a crystal-controlled B.F.O. for use when receiving C.W. signals, while the PV98A is designed for M.C.W. and R/T working. Both are extremely versatile and may be used on conjunction with Plessey or other manufacturers' units in commercial communication networks. In an F.S.K. installation, for example, the use of a PV98 in conjunction with stable oscillators PG81 and PG82, provides five-channel reception with the same high stability as the standard PVR80 equipment, and is hence suitable for unattended operation.

When the stability requirements are less critical,


One of the many printed circuit adaptations-a Belling Lee plug and socket combination.
the PV98 may be used with its own internal oscillators, making an F.S.K. installation available at very moderate cost. Operational trials have shown that extremely good performance can be achieved if the vernier controls are periodically adjusted.


One of the new " $C$ " core potted components-by' Whiteley Electrical.

Both pre-selectors comprise an R.F. stage, a crystal-controlled H.F. oscillator and a mixer, giving a low impedance coaxial output at a fixed frequency of $2.1 \mathrm{Mc} / \mathrm{s}$. The R.F. stage is tuned by pre-set circuits and the oscillator stage crystal controlled, the appropriate circuits being selected by a single five-position rotary switch manually operated from the front panel. Alternatively, the selector switch may be motor-driven from the rear, thus permitting remote control.

Additional facilities are available on the PV98 for operation from external oscillators. The tuned circuits are mounted as plug-in units, six ranges being provided to cover the frequency band.

Two pre-selectors and two receivers may be connected to form a dual-diversity combination, the receiver outputs feeding a conventional path selector or diversity switch unit. Under these conditions the channel frequency crystals need only be fitted in one of the pre-selectors, as the two may be operated with H.F. oscillator circuits cross-coupled, and similarly for diversity telegraphy two B.F.O. outputs are provided for the operation of both receivers


Plessey's new 3in. loudspeaker.
driven by a single crystal fitted in one only of the pre-selectors.

## Switches

F.M. slider switches were seen, and are intended for a variety of purposes, including F.M. radio and TV band switching. They utilise H-type contacts in groups of three, and as many as six groups may be incorporated on one switch. Being flat, the switches may be mounted horizontally or vertically and thus occupy the minimum of space. They are slideroperated, either directly or by some external means such as a cam or a simple rotary switch. Both springloaded and non-spring-loaded types were shown.

## Condensers

Gear-driven and in-built ganged capacitors were also seen. The integral ball drive variable capacitors types " $U$ " and " $S$ " are fitted with integral reduction drives having ratios of approximately $6: I$, thus simplifying the design of receivers, since additional cord or other slow-motion drives are unnecessary.

Control knobs shown are of various types, while the capacitors displayed included minjature plasticcased electrolytics, NATO-type electrolytics and tubular paper capacitors.

New sealed lampholders, type X.L.1, designed primarily for use on radio and electronic equipment, and holding lamps with standard miniature Edison

screw caps were shown. For these lampholders two interchangeable caps are available. One is a translucent moulding in a wide variety of colours to suit individual requirements. The second is a dimmer cap consisting of a light alloy die-cast fitting with a plastic filter coupled to a shutter.

Plessey have recently designed additional types of output transformers. Among those displayed were a miniature transformer for use in "personal" sets, in conjunction with sub-miniature output valves and where the primary direct current is limited to 3 milliamperes maximum : size 10 for use in battery portable and "personal" sets in conjunction with sub-miniature valves operating above 4 milliamperes direct current in the primary; size 13 a for use with 10 in . speakers operating in the 4 - to 5 -watt range, and size 15 for use with high-fidelity equipment, with a power output rating of 6 watts maximum.

## Loudspeaker

The 3 in . inset loudspeaker, shown on the lef? was specially developed in conjunction with S.R.D.E. for use in Services equipment and is suitable for
any application in which a small size, clear reproduction of speech and the ability to withstand extreme climatic conditions are essential requirements.

The inset loudspeaker is mounted in a sealed metal

backing plate by means of vibration-proof rubber mouldings, the connections being brought out of the case through ceramic seals. The front is protected by a perforated steel cover. The magnet system is of the totally enclosed centre-pole type. The frequency response has been specially adjusted to give maximum intelligibility of speech under conditions of high ambient noise and has extreme sensitivity over the range $800 \mathrm{c} / \mathrm{s}$ to $5,000 \mathrm{c} / \mathrm{s}$ with a frequency response extending to $400 \mathrm{c} / \mathrm{s}$. It will handle 1 watt and has a speech coil impedance of 10 ohms. The commercial versions are made with either round or square mounting flange.

## Tape Recorders

Among the several tape recorders shown were the special Ferrograph instruments shown by Messrs.

Wright and Weaire. These included instruments having tape speeds from $11 / 16$ c.p.s. up to 15 c.p.s. in a variety of portable and rack-mounted forms to meet the varying needs of industry, research, education and entertainment.

Auxiliary apparatus for use with Ferrograph recorders including:

Endless loop cassettes. Signal operation switch units. Pulse units.

Mention must also be made in this connection of the novel Bib tape splicer shown by Multicore. This has previously been dealt with in these pages, but has now been redesigned and is now known as the Mark 2 and incorporates two tape retaining clamps which, in addition to having extensions on them providing an even casier releasing arrangement, are also fitted so that the clamp openings both operate identically towards the tape recorder. It is claimed that this makes tape splicing a quicker and simpler job without losing any of the virtually foolproof advantages of the original Bib product.

Multicore Solders, Ltd., state that if an earlier model has already been bought and the user requires this latest modification, the work will be undertaken on receipt of the Splicer, properly labelled with the owner's name and address, and a postal order for 2 s .

## Potted Chokes

Another type of component which is coming into increasing use is the "C" core potted transformer and choke. Illustrative of these is the W/B component shown on page 335. This has a "C" core moulded in Aroldite, and is fitted with waterproof plugs and sockets. Other W/B exhibits included the complete range of Stentorian high fidelity loudspeakers, now incorporating the patented Cambric cone from $2 \frac{1}{2} \mathrm{in}$. to 18 in . in diameter, including 8 in. , 9 in . and 10 in . models fitted with universal impedance speech coils, providing instant matching at $3,7.5$ and 15 ohms.

## Millimicro-second Photography

THE use of image converter tubes as the shutters of ultra high speed cameras is now well established. The image of the object to be photographed is focused on the sensitive photocathode of the tube. When operating voltages are applied a duplicate image appears at the opposite end of the tube. To photograph fast transient phenomena the tube is switched on by voltage pulses synchronised with the phenomena, the actual photographic record being made by an ordinary camera.

In the past circuit limitations have fixed the lower limit of exposure at about $30 \times 100^{2}$ seconds. In a new Mullard instrument the use of a coaxial pulsing system enables the exposure time to be reduced ten times, i.e., to $3 \times 10^{-8}$ seconds.

In a demonstration at the Physical Society's Exhibition recently a voltage pulse produced by a very fast spark discharge travels down a coaxial line to trigger on the tube. At the same time the light generated by the spark is guided by mirrors to the photocathode of the image tube. When the light path is lengthened by moving the mirrors a point is reached where the image of the spark on the viewing screen disappears, because the light arrives at the photocathode after the voltage pulse which triggers the tube has died away.

## New Reflectometer System

ONE of the most important elements in the HewlettPackard reflectometer system for wide-range microwave impedance measurements is the new 'hp' 416A Ratio Meter. This instrument automatically combines forward and reverse signals and displays their ratio directly. Reflection coefficient is read directly on a front panel meter. A separate output terminal is provided to operate an oscilloscope or recorder. The Ratio Meter includes an R.F. power monitor indicating proper power level. Accuracy obtainable for swept frequency measurement is $\pm 0.015$ reflection coefficient ; for single frequency measurement, $\pm 0.005$ reflection coefficient. Model 416A may also be used to measure S.W.R. in connection with slotted lines. A reference voltage from the system power source applied to the ratio meter eliminates error due to amplitude variation.

Another important instrument in the reflectometer system is the new 'hp' 670HM Swept Frequency Oscillator. This equipment operates over a frequency range from 7 to 10 KMC . It may be manually tuned or motor driven to sweep any portion of this frequency band automatically. Sweep is at a velocity which is constant and sufficient to ensure a clear trace on a long-persistence oscilloscope. The instrument has a direct-reading frequency dial.

BI-METAL strips are fairly easy to obtain from surplus dealers, or they could possibly be made at home. Manufactured strips are capable of giving delays up to three or four minutes. When used in the circuit of Fig. 20, the double action ensures that the strip is always cold and ready for re-use as soon as it has completed its cycle.
It will be noticed also that, in the event of a mains failure, the energising voltage to relays $\frac{A}{3}$ and $\frac{B}{1}$ ceases. These relays would then de-energise and, on resumption of the mains supply, it would be necessary for the bi-metal strip to operate before H.T. could be obtained once more. The circuit, therefors, protects the equipment in which it is used against all foreseeable accidents. (The energising voltage for the relays should, of course, be derived "after" the on-off switch of the equipment.)

In last month's article we discussed a relay circuit by means of which it was possible to control three remote circuits over two lines (see Fig. 15 (b)). This month we shall examine a system which is capable of controlling many more operations whilst still requiring only the two interconnecting lines. As this system lends itself very easily to control by radio, this aspect will be given due prominence as well.

## The Uniselector

One of the most practicable devices for enabling a large number of remote circuits to be operated over a single pair of lines is the uniselector. (The uniselector


Fig. 22.-A simplified diagram illustrating the construction and operation of a forward-acting uniselector.

## the fourth article in a new series

 DEALING WITH A MOST USEFUL TYPE OF RADIO ACCESSORYBy J. R. Davies

(Contimued from page 305, May issue.)
is used in automatic telephone exchanges to interpret the impulses sent by dialling.)

A simplified diagram of a standard uniselector is shown in Fig. 22. Its operation is quite simple and is almost self-explanatory. When its coil is energised the armature is attracted to the core, causing the pawl to pyll the ratchet wheel through the distance occupied by one tooth and thereby move the wipers (i.e., the moving contacts at the ends of the contact or "wiper arms") to the next contact in the bank. On de-energising the coil the armature restoring spring brings the armature and pawl back to their original positions. The pawl, on returning, engages with the next tooth of the ratchet wheel. The detent spring (a fixed spring which bears against the teeth of the ratchet wheel), prevents the wheel from rotating backwards as the pawl returns.
It will be readily seen that, by energising and de-energising the uniselector coil the requisite number of times, it is possible to set the uniselector wipers to any desired position. With a uniselector, therefore, a large number of different operations can be controlled over two interconnecting leads alone, simply by sending the requisite number of impulses.

## Practical Details

The uniselector shown in Fig. 22 has 25 fixed contacts on the contact bank. This is a standard number (although not by any means necessarily the number used for all uniselectors), and corresponds to what is necessitated by a 10 -number telephone dial.

In practical use the 25 contacts shown in the diagram resolve themselves down to 24 , since, when one wiper is resting on the 25 th fixed contact, the


Fig. 23 (a).-A pair of bridging wipers. They are depicted here moving from one fixed contact to the next. (b) A pair of non-bridging wipers.
wiper on the opposite arm is resting on the first. The remaining 24 contacts then split into two halves of 12 contacts each, the first 12 contacts being paralleled with the remaining 12 contacts (i.e., contact 1 to contact 13,2 to 14,3 to 15 , and so on). With this method of connection it is unimportant whether the rotating wipers start a dialling cycle at contact I or contact 13. The 12 contacts" remaining " are the number required for dialling purposes. - Of these, one contact is used for "homing " or reference purposes, another is spare and the remaining 10 correspond to the 10 numbers on the telephone dial.

Although only one contact bank was shown in the simplified diagram in Fig. 22, uniselectors usually have at least two banks and often more.

Fig. 22 also showed two interruptor contacts ; these being operated by the armature and breaking. as the armature moved to the core. These contacts are used "for " homing "-a process" which will be explained later. It can be readily imagined that, if these contacts were connected between the coil and an energising voltage, they could cause the uniselector armature to vibrate in the same manner as does that of a bell or buzzer. The pawl, being fixed to the armature, would move in a similar fashion and would cause the ratchet wheel to be turned through one tooth for each. vibration of the armature; with the result that the wipers would rotate at a relatively high speed. $\because:$ :

Another point of importance lies in the fact that two different types of wiper may be employed. These two types are known as bridging and non-bridging wipers, and are illustrated in Fig. 23. Bridging wipers are used when it is necessary to ensure that no break occurs between contacts.

Returning to Fig. 22 again, it may be remembered that the pawl of the uniselector shown in this diagram moved the: ratchet wheel when that armature was attracted to the core. The force driving the ratchet wheel was derived therefore from the movement of the armature. In a large number of uniselectors the ratchet wheel is moved when the armature releases, the force driving the wheel being derived this time from the armature restoring spring. When the armature of such a uniselector moves to the energised position it causes the pawl to engage with the next ratchet wheel tooth. This latter type of relay is known as "reverse-acting." That of Fig. 22 is "forwardacting."

Having gone thus far into the theory and design of the uniselector, it would now be of interest to show how we can use it in a practical circuit.

## Circuit Design

Such a circuit is shown in Fig. 24. For this circuit we need a reverse-acting uniselector with two contact banks ; one having bridging wipers and one having non-bridging wipers. Twelve contacts are shown in the diagram for each bank, the remaining 13 (assuming a 25 -contact uniselector), being omitted for simplicity. The purpose of the circuit is to so control the uniselector that it will automatically select one of 10 circuits; control being effected over two lines. (It could also be controlled by a radio link; Sl and $\mathbf{S} 2$ keying the transmitter carrier, and a relay in the receiver closing their circuit on reception of the carrier.)

Although the circuit of Fig. 24 may be, at first sight, a little complex, this is only because it takes advantage of the refinements offered by the uniselector. The circuit could very conveniently be operated by a dial,
but, as such components are difficult for the experimenter to obtain, the switch and push-button are used instead.

## Operating the Uniselector

Let us now see how the circuit operates. We start off by closing SI. This energises relay $\frac{A}{3}$ and, after a second or two, relay $\frac{B}{2}$ as well (the latter being slow to operate and release). We then press the push-button, S2, a predetermined number of times corresponding to the number of impulses we wish to transmit to the uniselector coil. Each time S2 is pressed, relay $\frac{A}{3}$ de-energises, causing its break contact Al to close. Relay' $\frac{B}{2}$ does not de-energise during this period because it is slow to operate and relcase. Thus, each time relay $\frac{A}{3}$ de-energises, a circuit is completed from the positive source of supply via contacts B1 (closed all the time) and A1 (closed when relay $\frac{A}{3}$ de-energises) to the uniselector coil, and thence to earth. The uniselector armature then energises and de-energises (once for each pressure on the push-button), and correspondingly causes the ratchet wheel to be.turned through the requisite number of teeth. When S2 has been pressed and released for the last time it is left alone, whereupon relay $\frac{A}{3}$ stays energised and the uniselector comes to rest, its wipers positioned at the contact selected.

During this time, another cirçuit has come into operation on Bank 1. As soon'as we started to press the push-button, and the bridging wipers of Bank 1 left contact 1 of that bank, they commenced to complete (via contacts 2 to ${ }^{2} 12$ inclüsive), ame energising circuit to reläy $\frac{\mathrm{C}}{1}$; this circutit"being completed by the make contact. A2. However, due to the fact that relay $\frac{\mathrm{A}}{3}$ is centinually energising and de-energising whilst the push-button impulses are being sent, this energising circuit is intermittent. Relay $\frac{C}{1}$ is slow to operate and release and this intermittent energising current is. insufficient to cause it to energise. This condition would automatically be satisfied if a dial were used instead of SI and S2. With the push-button it would probably be necessary to send the impulses fairly quickly, and to try to keep the time of each approximately equal to the time between impulses. When impulse sending has been completed, and S2 is left closed, relay $\frac{A}{3}$ remains energised; and the energising current to relay $\frac{C}{I}$ becomes constant. Relay $\frac{\mathrm{C}}{1}$ therefore energises.
$\dot{\text { As soon as it }}$ it energises, a circuit is made, via contact $C 1$, contact $A 3$, and the wiper of Bank 2 , to the appropriate external circuit selected. It will be seen that the purpose of relay $\frac{C}{l}$ is to prevent unwanted
(Continued on page 341)
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circuits being operated until the uniselector has come to rest.

When the operation selected by the push-button and the wiper of Bank 2 has been completed, the uniselector can be brought back to its original state again and be made ready for the next cycle of events. This is carried out by the simple process of opening SI.

## Homing

The act of opening $S i$ first of all causes relay $\frac{A}{3}$ to be de-encrgised, whereupon its contact A3 breaks the controlled circuit selected on Bank 2. Another contact, A2, then breaks the energising circuit to relay C which, after a pause duc to its slugging, de-energises; and thus becomes ready for the next cycle.

Also, when SI opened and relay $\frac{A}{3}$ de-energised, a further break contact, AI, came into operation, causing the uniselector coil to be energised. (As the uniselector is reverse-acting this merely causes, the pawl to engage with the next tooth of the ratchet wheel). In addition, opening S1 also de-energised relay $\frac{B}{2}$, the armature of which, after a pause due to the slugging of the relay, releases. Contact BI then breaks, removing the direct supply to the uniselector coil ; whilst contact B2 makes, causing an energising circuit to be made to the same coil via its interruptor contacts and the bridging wipers of Bank 1. The armature of the uniselector then functions in the same manner as a buzzer and rotates the wipers. When the wipers of Bank 1 reach contact 1, the supply to the interruptor contacts breaks, and the uniselector comes to rest in its original position. (This last operation is known as "homing.")

The circuit is then ready for the next cycle of control.

## Practical Points

As may be imagined, the circuit of Fig. 24 can prave to be extremely useful, since it offers a convenient means of control. (There is, of course, no reason why the ten operations used in the circuit could not be increased.) The process of obtaining a uniselector should not be too difficult as these are beginning to appear on the surplus narket.

It was mentioned above that the uniselector could be operated by a radio link as well as by direct lines. Its employment in radio-controlled models appears, therefore, to be very attractive. Unfortunately, however, the uniselector is rather a heavy item of equipment and, in addition, it needs a fairly large energising current. It would be impracticable to energise it from any source of supply other than an accumulator. The use of the uniselector is therefore restricted, so far as radio control is concerned, to the larger and heavier models; or to static apparatus.

If the constructor wishes to overcome these difficulties, there is little to prevent him from using a home-made uniselector in place of the manufactured article. With care and ingenuity, it should be possible to make a simplified version which would be small, light, and moderate in current consumption. The "homing" circuit of the larger model could be dispensed with : and the entire operation could be reduced to that of a simple-step-by-step switch. It would probably be found easier to make the homemade model forward-acting, in which case it could be
actuated by sending single positive pulses from the transmitter. Without the "homing" circuit the uniselector would, of course, maintain the last position selected for it.


Fig. 24.-A uniselector control circuit. Its action is described in the text.

There are several small disadvantages to this simplified form of control, the most obvious being that the uniselector would operate every circuit through which its wipers passed. If it was considered to be of sufficient importance, this trouble could be obviated by using a relay which is slow to operate (see Fig. 11 (d)), it being energised by the same receiver circuit that operates the uniselector coil. Should a serics of controlling impulses be sent to the uniselector, this relay would remain de-energised. If, however, on the last impulse the transmitter key were kept closed, the relay would then energise and complete the selected circuit. On opening the transmitter key the relay would at once de-energise and the controlled circuit would be disconnected.
A second disadvantage of the simplified homemade uniselector lies in the fact that the operator has to remember at what contact he last left the uniselector wipers before he can send the requisite number of impulses needed to give a new order. This disadvantage could possibly be overcome by using another uniselector at the transmitter, this also being operated by the control button (or key), and serving to light appropriate indicator lamps. If the uniselectors in the model and the transmitter commenced operating in step, the indicator lamps would then show the state of affairs existing in the model.
(To be continued.)


WITH the meter connected as in position (b) the result would not be quite so accurate as the reactance of C , and C 2 would be bound to have some effect on the voltmeter reading-probably causing it to read something less than 50 volts.

Now let us suppose that the meter is connected as in position (c), and that the loudspeaker loading at this point is, say, 4 ohms, then with the valve delivering 1 watt of A.F. we should get a reading of something like 2 volts, by altering the " watts formula " round like this: Volts equal the square-root of watts times the resistance in ohms.

When taking power output measurements it is desirable to disconnect the loudspeaker and connect in place an equivalent value resistor capable of dissipating the full A.F. power.

Receiver alignment and overall sensitivity tests usually call for some means of measuring A.F. power, and such tests will be dealt with in full later on in this series; at this point, however, it is as well to realise that the universal multimeter can be admirably employed for this function.

## The Multimeter as an R.F. Indicator (18)

As is very well known, the detector circuit of a broadcast or television receiver is essentially an R.F. rectifier. The modulated R.F. signal is usually applied direct to the anode of a diode valve, or to the equivalent terminal on a crystal diode, and the modulated signal, together with a D.C. potential corresponding to the amplitude of the R.F., is developed across the associated load resistor.

Within certain limits, and excluding the stabilising action of the A.V.C. system, any variation of directcurrent in the load resistor will be governed by the amplitude of the R.F. voltage applied to the receiver. Advantage is sometimes taken of this factor as a broadcast receiver alignment aid, and is nearly always used to assist in the alignment of television receivers.

A typical vision detector circuit is depicted at Fig 21, in which the $5,600 \mathrm{ohm}$ resistor RI constitutes the detector load. To achieve signal indication in the vision channel, therefore, it is necessary simply to unsolder R1 from the chassis (earth line), and connect in series a D.C. milliameter having a f.s.d. of not more than 1 milliampere. As is shown in the circuit, it is desirable to decouple the meter with a capacitor Cl to avoid any possibility of instability -Cl should have a value in the region of $0.05 \mu \mathrm{~F}$.

It is readily possible by this means to observe with a high degree of accuracy any alteration in sensitivity of the vision channel at a spot frequency selected on the signal generator-which, incidently, should be unmodulated-as the result of adjustments to the tuned circuits. Normally a current in the region of

250 microamps should be expected to correspond to a peak-white picture signal.

As an alternative method-but one which is liable to induce instability under certain critical conditions -the actual voltage appearing across the diode load can be measured and used as a signal indication by connecting a multimeter set on the 0-1 volt range across the resistor-connécting the positive lead to the cathode side of the resistor.

If a multimeter of mediocre sensitivity only is available, such as one possessing, say, 500 ohms per volt, this can be used on the $0-20$ milliampere range and inserted in series with the anode load resistor of the video amplifier valve, being suitably by-passed by a capacitor-see Fig. 22.

The sync pulses in certain receivers are extracted from the cathode circuit of the video amplifier valve. The Murphy V114 series receivers use this arrangement, which is illustrated in Fig. 23. A stable and reliable method for signal indication in the vision channel on such a circuit is by measuring the volts drop across the cathode load resistor. A multimeter adjusted to the 100 volts range is suitable for this purpose.

If a very sensitive multimeter or valve-voltmeter is available it can be connected between the cathode of the picture-tube (assuming cathode modulation)


Fig. 21.-Showing how the current in the detector load resistor can be used to provide an indication of R.F.
and receiver chassis. The meter should be adjusted approximately for f.s.d. on a volts range under conditions of no signal. Since the tube is cathode modulated, the presence of a signal will cause a fall in potential at the anode of the video amplifier valve (relative to chassis), being the reverse of the previously described methods, where a RISE in voltage or current indicates the presence of signal in the vision channel.

It must be stressed that in all these tests it is most important to kesp the meter connecting leads as short as possible, and this applies especially where the meter lead is connected to the picture-tube cathode direct. As previously intimated it is not necessary to modulate the R.F. signal applied to the receiver under test, for we are not bothered about extracting the modulation content as a means of R.F. signal indication, we are simply measuring the rectified R.F. voltage to determine the magnitude of signaland thus the relative sensitivity of the R.F./I.F. circuits-arriving at the detector valve.

We are readily able to follow the rectified R.F. in D.C. form past the detector load resistor, and right up to the picture-tube, so far as television receiver is concerned. Because, as we are well aware, the post detectơr circuits are all coupled to retain the D.C. component of the vision signal.

With an ordinary broadcast receiver this does not apply, of course. Here we can measure the D.C. potentia corresponding to the R.F. only at the detector load resistor-past this point the D.C. content of the signal is lost owing to the isolating feature of the A.F. coupling capacitor.

## General Testing (19)

There is little doubt that a good multimeter represents the most essential tool of the experimenter and practical radio man. It should be the aim of all experimenters and home-constructors either to make or purchase an instrument having a sensitivity of, at least, 10,000 ohms per volt on the D.C. ranges. Although instruments of lower sensitivity possess a wide range of application, something just that little bit better is usually demanded for TV work where extremely high resistance networks are often encountered-indeed, it is when working on such circuits that one realises the limitations imposed by instruments of low and mediocre sensitivity.
The writer ventures to suggest that practically any fault in a broadcast or television receiver can be brought to light by knowledgeable application of a high sensitivity multimeter. This is not to suggest that other test instruments are superfluous to the general cause, indeed, this is far from true as we shall later realise in subsequent articles in this series. Certain critical adjustments demand the employment of instruments of an entircly different kind, and in the service department, where time is at a premium, a host of instruments is often desirable to reduce the time spent on a repair, and consequently keep its cost down to a respectable level !

Now to get back to the multimeter, we have so far covered its general applications-other tesis will, of course, suggest themselves to the experimenter from what has already been described.

Seemingly, voltage tests would appear to represent the main mode of analysing a faulty receiver. In the writer's opinion, this is not true, for it would be suggested that the whole of a receiver can be readily and conclusively analysed by current measurements.

The function of each valve stage can be individually analysed in this way without seriously disturbing the voltage distribution of the circuit. Defects ranging from the local oscillator to the output section can be pin-pointed simply by breaking the appropriate circuit and introducing a milliammeter.

As an illustration in this respect, it may or may not be realised that a milliammeter inserted in series with the grid resistor of the local oscillator valve will immediately indicate whether or not the valve is doing its job. The 0-1 milliampere range should be used for this test, and one can rest assured that the valve is oscillating if the grid current reading can be reduced to zero by shunting the tuned circuil with an $0.1 \mu \mathrm{~F}$ capacitor.

A less sensitive milliammeter can be inserted in series with the anode of the oscillator valve if more convenient, when an appreciable alteration in valve current by by-passing the tuned circuit with a capacitor will indicate oscillation. Any valve stage can be checked for spurious oscillation by adopting this mode of current analysing.

Excessive cathode current in the sound output valve accompanied by distortion of reproduction represents an example of current analysing that -would almost certainly lead the operator to suspe: trouble in the valve biasing network. A fautt of this nature is frequently caused by a coupling capacitor developing a slight leak, thereby permitting the grid to go positive and as a consequence outweighing its normal negative potential. This would be quickly proved as fact if, on disconnecting the capacitor from the valve grid, the cathode current reduced to normal.

As will have been noticed, this individual current test proved not only that the excessive current was in some way tied up with the accompanying distortion, but it also proved that the coupling capacitor was faulty without the necessity of testing it subjectively.

A short-circuit occurring in the cathode by-pass capacitor would undoubtedly provoke a similar set


Fig. 22.-With a less sensitive instrumemt R.F. is best detected by connecting an 0-20 mA meter in series with the video load resistor.
of symptoms, but here, again, removal of the suspect from the circuit would soon reveal its defect.

There is, of course, always the possibility of a faulty valve upsetting one's initial suspicions, although even a valve fault could more or less be eventually proved by performing a simple current test.

This same mode of testing can be readily adopted in any other stage. Erratic and intermittent faults may be found to require a lengthy current test, but there is no doubt about it, that somewhere in the circuit the current is going to alter as the result of the fault.

Faults in the A.V.C. system can be brought to light by introducing a milliammeter in series with the anode of each controlled variable-mu valve in turn. A substantial reduction in anode current is sufficient proof that A.V.C. bias is reaching the valves and reducing their gain in the proper manner when the receiver is tuned to a powerful station.

## Estimating Capacitance (20)

Some multimeters are separately scaled in capacitance values and embody features for connecting an add-on capacitance measuring unit. If one is not lucky enough to possess an instrument of this kind, however, a method of determining the approximate value of a capacitor is available.

The test boils down simply to that of measuring the alternating current passed by a capacitor of unknown value as the result of a known voltage. The capacitor and meter are connected in series across the A.C. mains-the mains voltage will be known from accurate measurement beforehand-the current in the capacitor may be read from the A.C. milliammeter, and the capacitance can be resolved with the aid of the following formula :

Capacitance in $\mu \mathrm{F}=1 \times 3184 / \mathrm{E}$,
where $\mathrm{I}=$ the current in amperes, and $\mathrm{E}=$ the applied 50 cps. voltage.


Fig. 23.-In certain video circuits the voltage variation across the cathode resistor provides a suitable R.F. indication.

Owing to the resistance of the meter being in series with the capacitor; a certain calculation error is inevitable ; nevertheless, this will:usually be so small as to be insignificant for normal purposes.

The main problem of a test of this kind is that of protecting the meter should the capacitor develop a short-circuit while being measured. Such a possibility can be completely avoided by introducing a resistor, as a current-limiting device, in series with the capacitor as shown in Fig. 24. The presence of this resistor will, unfortunately, provoke considerable error in capacitance calculation if the above formula is used, though as a compromise the resistor could be included initially and, then short-circuited during the shortest possible time that an accurate current reading can be taken.

Even so, this still leaves the meter as the target should anything happen to the capacitor, and since meters are expensive items and not readily replaceable, it is best always to leave a resistor in circuit to limit the current to a safe value, and introduce a factor


Fig. 24.- A method of determining the value of a paper capacitor-not suitable for electrolytic types. Fig. 25.-A method of determining the value of an unknown capacitor by comparing it against one of known value.
to the above formula to correct it for resistance. The following formula is quite suitable and should always be used when a resistance follows the capacitor :

Capacitance in $\mu \mathrm{F}=3184 / \sqrt{ }\left(\mathrm{E} / \mathrm{I}^{2}\right.$ minus $\mathrm{R}^{2}$, where $\mathrm{E}=$ the applied 50 cps . voltage, $\mathrm{I}=$ the alternating current in amperes, and $R=$ the resistance in ohms.

There is another interesting and fairly accurate method of measuring capacitance, whereby the reactarfice of a known capacitor can be compared against that of an unknown capacitance. A circuit set-up of such an arrangement is shown in Fig. 25, from which will be seen that the known and unknown capacitors are connected in series across an A.C. supply, and the voltage developed across the known one measured on a high-resistance A.C. voltmeter.

It will also be seen that if the leads connecting the unknown capacitor ( Cx ) to the circuit are shorted, the full A.C. supply voltage will be developed across the known capacitor ( C ), and the meter will read accordingly. It is often desirable to get this circuit working after the style of an ohmmeter, and for this purpose the variable resistor R permits the meter to be set to zero at full scale when the tests leads are shorted.
(To be continued.)


## The Business Names Act

APROPOS my recent remarks regarding. radio clubs, it is not generally known that in certain circumstances it is necessary for clubs to register under the Business Names Act, particularly where such clubs are run for profit. The provisions of that Act to some extent safeguard the members, for it is necessary to issue an annual balance sheet. The Act was designed to protect the public from exploitation by any person or persons who start a club with the idea of collecting annual subscriptions in return for some nebulous scrvice which only absorbs a very small part of the annual sum received. Suppose, to take as an example a hypothetical case, someone started a club such as the International Radio Guild, appointed himself as the proprictor, without providing anything in the rules for annual meetings or elections, membership being obtained merely by filling in a form and paying, say, 10 s. a'year subscription. Further suppose that the club caught on and obtained a membership of, say, half a mitlion. The annual revenue from subscriptions would therefore be $£ 250,000$. That is a considerable sum of money, and the members have every right to know what is happening to it and whether it is being spent in promoting the interests of the club. Clubs are of two types-proprictary clubs and members' clubs. The members of the latter are co-owners therein in law, but not so the members of a proprietary club who are merely the payingeguests of the proprietors. Nearly every properly run club is constituted, regulated and governed by rules which provide for the admission of members, the subscriptions payable, the conduct of the club, and for the retirement of the officers and expulsion of members.

In my view the I.S.W.L. is a proprietary club, and as it does not publish a batance sheet to each member. they have no means of knowing how the money is spent or how much is being received. No doubt the intention was that it should be a members' club. In view of the large amount of correspondence I have received, it seems desirable, if the League is to continue, that it should be reformed, remodelled and run on proper and democratic lines.

I referred last month to a letter sent out by the I.S.W.L. to all members. At least, I was informed that it had been sent to all members, but one member, who resides in Barrow on Humber, tells me that he did not receive it and the first he knew of it was when he read my notes. I gather that at one tinie the books of the league were in a bad state, and it took months to square them up. Many letters requesting subscription renewal were sent to members, and many sarcastic replies were received, with statements that they had paid up their subs. but had not received a receipt. Is it not wue that most of the members approached did not renew their subscription the following year? It seems obvious that an audit back to the days of the league's formation would now be impossible. Indeed, no audit could be possible, without a list of members.

I feel that the only satisfactory solution is for a Special General Meeting to be called with the object of appointing a sub-committee to investigate the club's affairs and issue a report thereon. This investigating committee should be given the widest possible powers of investigation.

## Commercial Radio ?

IF the I.T.A. commercial television programme proves successful will the advice given in the Beveridge Report be ignored and commercial radio programmes introduced? If such programmes are right for television they must be equally right for radio. The price of one minute of TV programme time is, I understand, just short of $£ 1,000$. I do not foresee that many advertisers are going to get their money back from such a transitory form of publicity; bearing in mind the many restrictions which are to be placed on the amount of advertising plugging which will be permitted. A cheaper form of visual advertisement would be through the cinema screens of this country, the coverage of which'must be at least equal to the TV viewing audience. In any case, it is obvious that commercial TV is designed to appeal only to very large and prosperous companies. for very few of the'smaller ones could afford $£ 1,000$ for a minute of programme time.

## The Newspaper Strike

O NE effect of the protracted newspaper strike was to force many more millions of listeners to switch on to the news programmes which they often ignored because they had already received the news from their morning or evening newspaper. It is a pity in some ways that the BBC did not more extensively enlarge its news features during the run of the strike, although I understand that a general agreement was reached amongst the publishers and the BBC not to do anything which would savour of an attempt to cash in on the absence of newspapers. But then the public was entirely cut off from the woild during the newspaper strike and had to rely upon the BBC news bulletins. There were some millions who could not listen in and had to rely upon gossip for their news. The results of the municipal elections were a matter of national importance and so was the resignation of Sir Winston Churchill. Large numbers of the public were unaware of the results of the elections.

As new's dissemination is a matter of national interest the Government itself should have produced a daily national news sheet as it did during the General Strike, if it was impracticable for the BBC temporarily to fill the breach. It is true that the BBC did add a few minutes to the news bulletin, but the extra time was totally inadequate. It is on such occasions that a very small pocket portable with single earphone would be of great advantage, especially if it were designed to run on very low H.T. voltage in order to keep the weight down. The quality of reproduction would be secondary in importance. If any readers have built such receivers I should like to have details of them.

# More About BBC Frequency Modulation 

LATEST NEWS CONCERNING THE V.H.F. STATIONS`

THE BBC has placed an order with Marconi's Wireless Telegraph Co., Ltd., for the construction of 26 V.H.F. frequency modulated (F.M.) transmitters for sound broadcasting. Delivery will commence within 14 monthss.

These transmitters, of Marconi design, comprising 24 of $4 \frac{l}{2}$ kilowatts power and two of 10 kilowatts, will form part of the BBC's plan to provide a powerful reinforcement to the coverage of its present mediumand long-wave stations, by the use of V.H.F. F.M. stations.
As envisaged, the $4 \frac{1}{2}$ kilowatt transmitters will operate in parallel pairs, each pair handling one programme. Thus, six of these transmitters will be used on each three-programme station. The two 10 kilowatt transmitters will be used in parallel at the BBC's existing V.H.F. station at Wrotham, where there are already two 25 -kilowatt Marconi transmitters.

Nine Marconi station monitors have also been ordered by the BBC for use on this project.

## Technical Details

Many new features are incorporated in the transmitters and the performance of each is outstanding in its own particular output-category.
These transmitters, comprising units of basically similar designi, but suitably arranged to provide the variou's ouitput ratings, operate on the frequency band 87.5-108 Mc/s.
Air-cooled valves are used throughout, leading to a simplification of equipnent, with reduced installation and mainteriance costs.
The F.M.Q. drive unit used on these transmitters, employs a frequency-modulated. quartz crystal and supplies an output at the carrier-frequency. This signal is then amplified to raise it to the required output tevel, the number of amplification stages used depending on the rated power-output of that particular type off transmitter.

The-initial.A.F. amplification stages consist of at double teitrode stage, capacity coupled 10 a pair of tetrode valves, the anode circuits of which take the form of quarter-wave balanced lines magnetically coupled to a 50 -ohms line, from which the output is taken sat 250 W .

A third amplifier stage is added (in this case a triode), for the 1 kW equipment. The input is fed via the 50 -ohms line to the cathode, the two cathode leads bcing at the same. R.F. potential and forming the inner member of a tuned unbalanced line. The grid is grouided and the anode circuit takes the form of a concentric line whose electrical length is variable by moving an R.F. short-circuit along the linc. A tuncd inductive coupling takes power from the anode line and feeds to the aerial termination.

For the $4!\mathrm{kW}$ transmitter a further identical triode amplifier stage is added; while for the 10 kW case yet another amplifier stage is used, employing two such arrangements in parallel.

## Northerin Transmitter

A further contract for the supply, installation and setting to work of six 10 kW V.H.F. frequencymodulated transmitters to serve the North of England
area has also been placed. The order also calls for the provision of the necessary driye, phasing and monitoring equipment, and four combining units.
The six transmitters for the North of England are to be sited at Holme Moss, and will féed inio the slotted aerial array which forms an integral part of the existing 750ft. television mast, in accordance with the BBC's plan for shared TV/FM sites. The transmitters will operate in parallel pairs, one pair handling the Home programme, another the Light, and the remaining pair the Third programme. Each unit will therefore have an output power of 20 kW , which, in conjunction with the high-gain aeriat, is estimated to produce an effective radiated power of 120 kW . It is anticipated that the new station may become operational in the latter half of 1956.
The range cannot as yet be stated with certainty, but the BBC estimate the coverage to be roughly over an area bounded on the north by a line running from Barrow to Bridlington and on the south by a line from Rhyl to Cleethorpes. There will, however, be local variations in the fringe areas because of the screening effect of hills and large buildings.


Assembly work in progress on a Marconi $4 \frac{1}{2} \mathrm{~kW}$ F.M. transmitter for the $B B C$.

## REFRESHER COURSE IN MATHEMATICS <br> $8 / 6$, by post $8 / 10$. <br> by F. J. CAMM

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| in. p.m. |  |  |  |
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$\begin{array}{lllrl}\text { Band } 5 & 3.7 & \text { to } & 7.5 & \mathrm{mc} / \mathrm{s} \\ \text { Band } 6 & 7.5 & \text { to } & 15.0 & \mathrm{me} / \mathrm{s}\end{array}$
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|  |  |  | H. | D. | Serewed Boss |  |  |
| * 32 | 350 | 400 | 27 | 1 | $\frac{1}{2}$ | 312 | 91. |
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## THE TELEGRAPH CONDENSER CO. LTD

RADIO DIVISION: NORTH


TTHE east coast receiving station of R.C.A. Communications Inc. is located at River Head Long Island, a few miles east of Rocky Point, and approximately 70 miles from the centre of New York City.

This site, which when purchased 30 years ago was barren land with sandy soil, is flat for many miles around. Four miles north is Long Island Sound, with the Atlantic Ocean eight miles to the south.

The original receiving station dates before the short-wave era when long waves were used exclusively.

## River Head To-day

The River Head site is divided into two areas, the east being 600 acres and the west 1,264 acres. The receivers as used at River Head are gigantic affairs when compared with short-wave receivers as we know them. Built-in shielded sections and rack mounted, a complete receiver takes up a considerable amount of space.

## The Receivers

The total number of short-wave receivers available is 69 diversity groups divided into three sets. It is interesting to know that the number of valves used and in operation total 10,000 .

When one considers the number of spare radio valves which must of necessity be kept for replacement purposes the grand total will be rather staggering.

## The M1005 Receiver

This is the type of receiver used at the station, and here are brief details concerning it. Taking all types of valves as listed, these total 78. This receiver has three R.F. amplifiers, $3-6 \mathrm{Mc} / \mathrm{s}, 6-12 \mathrm{Mc} / \mathrm{s}$, 12-24 Mc/s respectively, followed by three further R.F. amplifiers and three third R.F. amplifiers. All stages at stated Mc/s, excepting the following :

Then follow three detectors and three oscillators, a first I, F. amp. First Stage at $450 \mathrm{kc} / \mathrm{s}$ and a first l.F. Second Stage. Next comes converters $Z, Y$ and $X$, a $400 \mathrm{kc} / \mathrm{s}$ oscillator, isolation amps. $Z, Y$ and $X$, followed by three $50 \mathrm{kc} / \mathrm{s}$ second I.F. amplifier stages, two diode detectors, monitor oscillator, monitor detector, monitor


A corner of the main operating room at Broad Street, New York.
mission line used at River Head is $4,000 \mathrm{ft}$. This is a four-wire, and not a transposed line.

## Aerials

All aerials are directive; 24 Rhombic and 50 fishbone types. These are spaced in groups of three and arranged for space directivity reception 800 ft . apart. That this is an extensive aerial system will be appreciated, for there are 600 aerial poles which are 60 ft . to 150 ft . high. There are also eight stcel towers 160 ft . to 200 ft . high, together with 6,000 transmission line poles 16 ft . to 35 ft . high.

## H.F. Traffic

The H.F. reception section averages a daily peak of 52 signals. Leased land lines are used between River Head and New York, as follows :

Sixteen lines 50 to 8,000 cycles transmission bands to New York, and two lines via Rocky Point to New York. There are also six lines from New York to Rocky' Point. For multi-channel transmissions 18 filter groups are available.

## Radio Programmes

Another section deals with radio programme material, and Radiophoto transmissions. On this service, six diversity groups of three sets and one single-sideband receiver are used; three fishbone and 13 Rhombic aerials. All the aerials are directive.

## Frequency Checking

There are three measuring positions. They cover the following ranges : $10 \mathrm{kc} / \mathrm{s}$ to $500 \mathrm{kc} / \mathrm{s}$.

In conjunction with this department three M.F. aerials, seven H.F. aerials of the fishbone type and V.H.F./U.H.F doublet aerials are used. Primary crystal standards are accurate to 1.5 parts in 10,000,000.

## Comment

Readers of this article will appreciate that TransAtlantic and world-wide radiophone communication is radio in a very big way, far beyond what we carly listeners used to imagine in the days when we were, as it were, in on the early tests of commercial point-to-point and ship-to-shore-telephony.

During the years the author has contributed to this journal he has from time to time stressed the importance of an efficient aerial system, however simple its form may be, in order' to assure the maximum signal voltage transfer. That this is essential is proved when one realises that professional radio engineers like those of R.C.A. go to great expense in orcier to assure reliable communication day by day.

## Diversity Reception

This is carried out with a view to receiving the required transmission at a constant and satisfactory signal level. To this end three separate aerials are coupled to three separate receivers tuned to the one frequency. By this means fading effects are compensated for. As is well known when using aerials located some distance apart, fading at any one time is not equally pronounced in all. Thus if the combined inputs are fed into a common output a constant signal or nearly so will be obtained.

## New "Stereosonic"

at each end of one wall of the room, so that the subtended angle to the listener is approximately between 60 deg. and 90 deg. The new H.M.V. "Stereosonic" reproducer will give good resilts using any form of stereophonic tape, including those made by using spaced microphones, but the best results will be obtained with the new H.M.V. "Stereosonic" tapes made on the principles described above.

Using such tapes, a full field of sound is obtained, stretching across the space between the two loudspeakers. When the sound comes from a solo instrument intended to be in a central position, then a single apparent source of sound is heard from a position half-way between the two loudspeakers. The general enhancement of the realism of the reproduced sound is apparent at any position in the room, although the exact position of the sound source is reproduced with complete accuracy only over a listening area near the centre line between the two speakers.

An important feature of the "Stereosonic" records is that the listener can obtain an apparent increase of dynamic range. The reasons for this are not yet fully understood. The fortissimos for a given sound level (as measured by a sound level meter) appear to have a greater volume of sound compared to the normal single sound source, while the pianissimos can be reduced to a lower level with reasonable discrimination and clarity. A further feature is that a greater degree of reverberation is permissible on the recording than would be permissible with single-channel recording.


A completely steerable 30 ft. parabolic reflector used as a radio telescope at the Jodrell Bank experimental station.

AN input which is only a minute fraction of the noise power generated in the receiver still gives a clear indication, and with this improvement it may be well worth while to connect an automatic recording instrument instead of the meter to the output. This can even be calibrated directly in units of field strength.
An aerial-in order to be usefully employed in a radio telescope-must be highly directional. To a certain extent this property is inherent with all types


Fig. 1.-4-element stack of half-wave dipoles. Inphase feeding is achieved by crossing the feeders between clements.

## Making a Radio Tele

## A SIMPLE HOME-MADE BUT EFFICIENT INSTRUMENT FOR THE EXPERIMENTER

## PART 2.-AERIAL SYSTEMS

## By W. Schroeder

of aerials, but not nearly sufficiently for the purposes of radio astronomy.

Best known among the directional aerials, of course, is the usual television aerial, normally consisting of a half-wave dipole with a reflector placed behind it.

A simple dipole has a power gain of 1.64 over an aerial which receives at equal strength from any direction. Such a purely theoretical aerial is called an isotropic aerial and cannot be realised in practice. It is conceived merely as a standard for comparison.

For use in radio astronomy, however, acrials of considerably higher gain and directivity must be used. This can easily be achieved by using an array of several half-wave aerials and feeding them in phase (Fig. 1). Highest sensitivity is in the direction Z. The polar diagram of this four-element system is shown in Fig. 2. Sensitivity is zero at an angle of 30 degrees from the direction of the highest sensitivity (Z), and by increasing the angle a smaller maximum of sensitivity is found. These "side lobes" are present with all multi-element arrays.
The beam width of this aerial, which is reckoned to the angle at which half maximum sensitivity is measured ( -6 db ), is about 35 degrees. An eightelement system would have a beam width of 25 degrees, and by further increases the directivity can be made greater still.

If several such "stacks" are placed side by side, the directivity in the horizontal plane can be increased too. Such broadside arrays (Fig. 3) are used quite frequently in radio astronomy, and they are especially effective if a conducting screen, usually


Fig. 2.-Polar diagram of a system of feur halfwave dipoles. Highest sensitivity in direction $Z$, zero sensitivity in directions 30 deg. and 90 deg. ( $X$ ).
consisting of wire netting, is placed one eighth of a wavelength behind the elements. This doubles the gain of each half-wave aerial, and the combined gain of the system in Fig. 3 would be $2 \times 1.64 \times 20=$ 65.6.

The screen at the back should overlap the aerials by at least half a wavelength. The beam width of the array shown is about ${ }^{9} 30$ degrees in any plane.

A more economical method, though a little more


Fig. 3.-Graphical representation of a 20-element broadside array, with screen behind. Power gain about 66, beam width 30 deg.
difficult to construct, is the erection of an array of Yagi aerials. A single unit of this type, consisting of a folded dipole, a reflector 0.15 wavelength behind the dipole, and five directors, 0.434 wavelength long, and spaced at half-wavelength intervals in front of the dipole, has a gain of about 20 and a beam width of 35 degrees. The measurements given represent the optimum values found experimentally.

Five such systems can conveniently be mounted on a frame which should be as large as possible to keep the beam width small (Fig. 4). Working on a wave. length of about 1 metre, and the Yagi elements at the corners of a square with sides 20 ft . long, the arrangement will provide a gain of 100 and a beam width as small as 10 degrees. (Beam width is approximately 60 wavelengths divided by length of the sides of the square. This formula also applies to broadside arrays).

Both broadside arrays and Yagi aerials, although extensively used in radio astronomy, have one great disadvantage : they have to be constructed for a specific wavelengih, and any desired change of this requires a complete rebuilding of the aerial system.

## Overcoming a Disadvantage

This disadvantage can be overcome by the use of a parabolic reflector. This usually consists of a circular frame over which a wire mesh is stretched in such à manner that a crosssection through the centre is a true parabola, and the dipole is situated at the focus.

Theoretically such an aerial has a power gain of $\frac{4 \pi \mathrm{~A}}{\lambda^{2}}$ where $\mathrm{A}=$ area of the aperture, and $\lambda=$ wavelength. The spacing of the wire mesh introduces some losses, however. These amount to 20 per cent. if the wires


A partly adjustable broadside are spaced $\lambda / 10$ apart, and 35 per cent. if spaced $\lambda / 8$. The latter is the widest practicable spacing.

An aerial system of 20 ft . diameter, working on a wavelength of 1 metre, and with wires spaced 5in., would have a power gain of about 220, quite enough to feed our receiver and give useful indications of the radiations which reach us from the Milky Way. The beam width of the system would be in the region

of 10 degrees, quite a useful value for its intended application.

The construction of such an aerial is comparatively simple, especially if it is not required to be stecrable. This need be no great disadvantage, as a certain amount of steering is possible by tilting the arm which carries the dipole at the focus. Without much deterioration of gain or beam width a deflection of up to 15 degrees can be achieved.

## Aerial Construction

The framework can be constructed of metal or timber, the simplest method being a number of poles stuck in the ground. The height of the uprights must be one quarter of the diameter of the framework, and

of sereral humbed half-wave
as near parabolical as possible, 12 uprights at least are necessary, but for larger systems working on shorter wavelengths, this number must be increased. At no point must the shape of the finished wire mesh depart more than $\bar{i} / 8$ from the true shape of a paraboloid.

In Fig. 5 are shown the suggested measurements for such an aerial, showing two of

## Table 1

centre. Height from ground.
1 in.
3 in.
7 in.
12.5 in.
19.5 in.
$28 \mathrm{in}$.
38 in.
50 in.


A fixed array of Yagi aerials for the reception of extra-terrestrial radiations.
the dipole at the centre is fitted at the same height. To make the shape of the reffector
the uprights which are opposite one another with the arm carrying the dipole in the niddle, which it should be possible to tilt in a north-south dire ction. In the centre, a wire ring is fitted, which should


Fig. 4. - End-on view of an array of five Yagi-ariads. For $l=20 \mathrm{ft}$., and on a wavelength of 1 metre, the beam width is 10 des., and the power gain 100 .
have a diameter of 25 in ., and is fixed lin. from the ground. Twelve wires are soldered to this ring, and their other ends are laid over the tops of the uprights, and weighted down to keep them taut. At a distance of 25 in . from the centre, spacing wires are soldered to the twelve wires in such a manner as to keep the solder joints 3 in. from the ground. Further spacing wires are then added according to the measurements given in the table.
For smaller or larger aerials these measurements must be decreased or increased in proportion.

When all spacing wires are in place, the ends of the twelve mains wires are fixed to the tops of the uprights, and they should then all be in the shape of parabolas.
Starting from the innermost ring, further rings are soldered to the main wires at intervals of 4 in . Finally, another number of wires are soldered between the main ones, starting from that ring where the distance between the main wires just exceeds 4 in., and carrying on to the rim of the "bowl," so that at no point of the completed mesh there is a space of more than 4 in . between any two wires. A reflector built to these measurements can be used at wavelengths down to 70 cm . ( $420 \mathrm{Mc} / \mathrm{s}$ ) and at that frequency would have a power gain of about 300 . If the spacing of the mesh is reduced to 2 in ., the gain reaches nearly 500 .

The dipole at the focus is best fitted with a reflector," behind it, on the side away from the "bowl." Although this has little effect on the gain, it considerably reduces the side lobes of the system.

It is necessary to match the aerial fairly accurately to the receiver input, as any mismatch not only results in a loss of power, but also introduces additional noise in the aerial, as the lost power is used to produce a thermal noise, and as the strength of this would not be constant it would lead to inaccurate indications.
If the aerial is built on level ground the beam is directed towards the zenith, and the measurements of the receiver therefore relate to that part of the sky which is right overhead at the time of observation.

As the Milky Way at certain times during the day reaches this position, fairly strong signals will be recorded. The radio telescope can be kept running all day long, and if it is made self-recording, a graph will be obtained showing the strength of radiations reaching us from different parts of the sky. These measurements can be compared with those taken 'on other days, but on the whole it will be found that the field strength recorded will be the same for any particular part of the sky.

By tilting the supporting arm of the dipole, the beam is displaced, and another part of the sky can be investigated. In each position, the beam sweeps out another circular strip of the sky, about 10 degrees wide, as the earth makes one revolution with regard to the fixed stars. When the possibilities of the aerial have been exhausted, the change to another wavelength offers new scope.
Hydrogen gas, which is fairly common in outer space, emits under certain conditions a strong radiation on a wavelength of 21 cm . The recording of this radiation and of the Doppler-shift associated with it has led to the most important results achieved by radio astronomy. So far, no other such " spectrum lines" have been discovered, but they should, theoretically, be present on some longer wavelengths. Here is a field of research for the keen amateur.
As "new stars" and other surprises in astronomy are almost always discovered by amateurs working with quite small telescopes, it is not impossible that in radio astronomy unexpected changes or developments are to be also first noticed by an amateur, in spite of the watching eyes of the scientists of Manchester University, who, at the end of this year, will take into use a new radio telescope of 250 ft . diameter.

## Correction

In the first column on page 290 of last month's issue it was stated that the receiver would give certain results with an input of only 10-14. watts. This figure should, of course, have been $10^{-14}$ watts.

## News from the Clubs

CLIFTON AMATEUR RADIO SOCIETY
Hon. Sec.: C. H. Bullivant, G3DIC, 25, St. Fillans Road, Catford, S.E.6.
$T^{\text {HE first D.F. contest in the } 1955 \text { series takes place on Sunday, }}$ May 8, in the vicinity of Farnborough. Kent. The club station, G3GHNN/A, will be operating portable between 11.00 lirs, and 16.00 hrs . on a frequency of $3504 \mathrm{Kc} / \mathrm{s}$ and as only low power will be used reports and QSOs will be welcomed.

Tape recording will be covered by Mr. L. Barnes on May 20 and the evening will comprise a demonstration and talk on the equipment being used.

Meetings are held every Friday at the clubrooms. 225, New Cross Road, London, S.E.14, at 7.30 p.m. Details of menbership may be obtained from the Hon. Secretary.
SOUTHEND \& DISTRICT RADIO SOCIETY
Hon. Sec.: 3. H. Barrance. M.B.E. (G3BUJ), 49, Swanage Road. Southend-on-Sea, Essex.

## O

 N April 1 the judging of exhibits of home-built equipment for the Hudson and Pocock Cups took place. There were good number of entries. and the judges spent some time in determining the winners. Mr. J. L. Goss won the Pocock Cup. and Mr. A. D. Asher the Hudson Cup. These, together with a runner-up prize, won by Mr. C. W. Crags. A.M.I.E.E., were presented by Mrs. Goss. wife of the society's chairman. at the Hamfest, held at the Royal Stores Restaurant, the following day. Another Cup. the Peck Cup. was also prescated to Mr. day. Another Cup. he peabrook for the berformance during 1954, in "direcR. K. Seabrook for the best performance duringCOVENTRY AMATEUR RADIO SOCIETY
Hon. Sec. : J. H. Whitby. G3HDB, 24, Thornby Avenue, Kenilworth, Warwick shire.
FORTHCOMING Programme. At 9. Queens Road ( $7.30 \mathrm{p} . \mathrm{m}$.) May 9, "Receiver Servicing," G3HDP. May 23. "Fre" quency Modulation." G6WH. June 2, Night on the Air. June 4-5. National Field Day. June 6. Junk Sale. June 20, "V.H.F.". G3BAK. July 2-3, Field Day. July 4. Lecture. July 7, Night on the Air. July 18. Lecture. A ugust i. No meeting. August 4. Night on the Air. August ${ }^{\prime}$ 7. Field Day. September 1, Night on the Air. September 4, Field Day.

## EAST KENT RADIO SOCIETY

Hon. Sec.: D. Willianis, "Llandogo," Bridge, Canterbury, Kent.
THE society meets fortnightly on Tuesdays at "The Two - Brothers." Northgate Street, Canterbury. Nearly all radio and electronic subjects are covered. Raffles are held regtularly and lectures and demonstrations also. New members are joining nearly every meeting.

New members are welcome and visitors in the district.
TORBAY AMATEUR RADIO SOCIETY
Hon. Sec.: L. H. Webber, G3GDW, 43, Lime Tree Walk, Newton Abbot.
STEPS are being taken to assist our new blind member. Geo. Western, BRS. 20605 , to get his gear together, and to arrange additional help in technical and Morse instruction for him.

Volume Controls
Midgot litiswa long mpindles Guaran－ teed year．Al valuce
10,090 ohme to 2 Meg． lo，000
No SH．
No Sw．B．P．St．D．P．Sw coA工 PLUGS 4／0 COAX PLU LINE CONHECTO UTL CON BECTOR $1 / 2$ BALANCED TWIN FEEDER per yd．ed TWIN SCREENED FEEDER per yd．1／－$\}^{80}$ 50 OHM COAX CABLE 8d．，per 5 J ．॥in．lia． TRI界NERS，Ceramic， $30,70 \mathrm{pf}$. gd． 100 pt ． $150 \mathrm{pf} ., 1 / 3$ ； $950 \mathrm{pf} ., 1 / 6$ ；ti00 pf．， $1 / 9$. 4 R．，3d．；- All values： 10 ohms to 90 mei．．．
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 2／－eit．Also 15,000 ohirs－i33，000 ohms $\overline{\bar{a}} \mathrm{w} ., 1 / 9$ ； 15,040 ohms－ 33,000 whus 10 w．， $2 / 3$.
WIRE－WOUND POTS．3 WATE．FAMOUS MAKE． Pre－Set Min．T，V．Type．Standard size Po 2 is Knurled slotted Knob．Spindle．Iligh tiratle． All valpes＂J ohins to 30 K．，b／efa． 50 K．． $4 /=$ bitts ：$\overline{0}$（arben Track $\overline{0} 0 \mathrm{~K}, 4 \mathrm{t} \boldsymbol{2}$ SLeg．， $3 / \rightarrow$ all Values．Itef whms to $\overline{\bar{n}} \mathrm{~K} ., 5 / 6$ ； $100 \mathrm{~K} ., 6 / 8$. W／W EXT．SPEAKER CONTROL 10 の， $3 /-$ O／P．TRANSFORMERS． standard Pentomle 4／8 ditto tapped prim．， $4 / 9$ ；small pentosle， $3 / 9$

 LTNX，3h．v30 ma．，13／6，SIMPLEX，1t h．JJo man．， 10／6．HATMS TRANS，－Made in our own workshnps to high＇grade specification，Fully＇ibter－leaved and impregnated．Tapped prini． 500 v．+550 v．，Heaver

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TOGGLE SWITCRES EX－GOFT．－＂On－off，＂9d，
 3d．5d．；3i－per t th．T．c．uire，is to 28 e．w．g．， per yd．， $2 \mathrm{~d} . \mathrm{Pl}$＇．Connecting nire， 10 coleurs． single or strunied，2d．yd．
8LEEVING．－Various coluurs． 1 mm，and 2 inm ．． 2d．Yd． 3 mm ．and 4 nms．，3d．yal．is mm．Sd，yd． 4 B．A．or 6 B．A．Nuts，Bolts atud Waster Kits，$\frac{1}{4} \mathrm{D}$. sin．，or ilin． $1 /=$ per doz．
FUSES．－ $1 \frac{1}{2 i n}$ ．all values 60 ma ．to 10 a ． $4 d$.
ALADDIX FOR EERS and cores．tin．，8d．；fin．， $10 d$.
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> No. 1.-PYE 18A AND G18K SERIES
switch to be brought into operation will impair the receiver's apparent sensitivity, and may lead one to suspect a fault somewhere in the aerial coupling circuits.

A triode-hexode frequency changer valve (ECH35) produces local oscillations in the triode section, and receives the aerial signals on the first grid of the hexode section. The appropriate aerial and oscillator coils are brought into circuit by the waveband rotary switch, each contact on which is shown on the circuit, for simplicity, as a straightforward makebreak switch.

L4 is the short-wave coil and T1 the short-wave trimmer; L2 is the short-wave aerial coupling coil. L 5 is the medium-wave coil and L6 the long-wave coil. When an external or plate aerial is used, L3 acts as medium and long-wave aerial coupling coil. T2 is the medium-wave aerial trimmer. The selected coils being tuned by C3.

The oscillator grid and anode coils are selected in a similar manner; $L 7 / 10, L 8 / 11$ and $L 9 / 12$ being the short, medium and long-wave coils respectively. These are trimmed in the same order by means of T3, T4 and T5, and oscillator tuning is performed by C4 which is ganged to C3. A degree of fixed padding is achieved by the $570^{\circ} \mathrm{pF}$ capacitor C5.

Mixing takes place within the valve V1, and a 465 $\mathrm{kc} / \mathrm{s}$ intermediate-frequency signal occurs across L13. This is conveyed, via the first I.F. transformer (I.F.T.1), to the signal grid of the I.F. amplifier valve
(Conimued on page 361)


Fig. 1.-Circuit diagram of Pye 18A, etc. The circuit is drawn with the waveband switch in the long wave position, and the tone control switch in the "fidelity" radio position. Switch $S 1$ is shown in the open position as wacitor values in microfarads, unless otherwise stated.

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A MPSH-PCIIE, 3-4 watt IfIGTI-A AIN AMPI.IFIER FGIR 83/\%6. FOr mains input 200-250 v. $50 \mathrm{c} / \mathrm{s}$. Complete kit of parts including circuit. point to point Wiring diagram, and instructions. Amplifler can be used with any type of Feeder Untt or Pick-up. This is not A.C.ID.C. $400-0-400 \mathrm{v}$. trans. (Output is for $2-3 \mathrm{ohm}$ speaker.) Supolied ready for for $2-3$ ohm extra. Carr. 2.6. Descriptive leaflet, 7d.

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$300-0-300$ v. $100 \mathrm{~mA}, 6.3$ v. 4 a, 5 v. 3 a $.22 / 9$ $350-0-350 \mathrm{v} .100 \mathrm{~mA}, 6.3$ v. 4 a, 5 v. 5.3 а ... 22/9 $350-0-350 \mathrm{v} .100 \mathrm{~mA}, 6.3 \mathrm{v} .4$ a. C. $\mathrm{T} .22 / 9$ $0-4-5 \vee .3$ a
$350-0-350$ v. $150 \mathrm{mA}$..6 .3 v.. 4 a. 5 v. 3 a $\ldots 29 / 9$ $350-0-350$ v. $150 \mathrm{~mA}, 6.3 \mathrm{v} .2$ a. 6.3 v. 2 a . $29 / 9$


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C.T. 0-4-5 v. 3 a ...
$250-0-250$ v. 100 mA .6 .3 v. 6 a, $\dddot{5}$ v. 3 ä $27 / 9$
$250-0-250$ v. 100 mA .6 .3 v. $6 \mathrm{a}, 5 \mathrm{v} .3$ a.
for R135j conversion .... $300-0-300$ v. $100 \mathrm{~mA}, 6.3^{\prime}$ v. -4 v. 4 ä,
$\begin{array}{ccccc}\text { С.'. 0-4-5 v. } 3 \text { a } & \cdots & \cdots & \cdots & \cdots\end{array}$

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1.1 a for V'CR97, VCR.517,

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| 17 | 1/6 | 2/4 | 1/6 | 2/4 |
| 18 | $1 / 6$ | 2/5 | 1/6 | 2/5 |
| 19 | 1/6 | $2 / 6$ |  |  |
| 20 | 1/7 | $2 / 7$ | 1/7 | 2/7 |
| 21 | $1 / 7$ | 2/8 | 1/8 | 2/8 |
| 22 | 1/8 | 2/9 | 1/8 | 2/9 |
| 23 | 1/9 | 2/10 | 1/9 | 2/10 |
| 24 | 1/9 | $2 / 11$ | 1/9 | 2/11 |
| 25 | 1/10 | 3/- | 1/10 | 3/- |
| 26 | $1 / 10$ | 3/1 | 1/10 | 3/1 |
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| 33 | 2/2 | 3/8 | 2/4 | 4/1 |
| 34 | 2/2 | 3/9 | 2/5 | 4/3 |
| 35 | 2/3 | $3 / 10$ | 2/6 | $4 / 5$ |
| 36 | 2/3 | $3 / 11$ | 2/8 | 4/8 |
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V 2 (EF39). On the radio position the function of this valve is quite conventional, the "amplified I.F: signal being developed in the second I.F. transformer (I.F.T.2), and passed on for demodulation to the signal diode in V3 (EBL31): The filtered A.F. signal thus appears across the 1 megohm volume control R2, and is taken through the grid-stopper R1 to the signal grid of the pentode section of V3.
This performs as the sound output valve and converts the signal into power, in conjunction with the output transformer TI, to drive the loudspeaker (L.S.)

Some of the I.F. signal voltage is taken by way of C6 to the A.V.C. diode in V3. This signal is thus rectified, loaded by R3 and filtered by R4 and C6, and used as A.V.C. bias for V1 and V2. A delay voltage and bias for $V 3$ pentode section is given by the two resistors in the cathode circuit of V3.
The network comprising switches, resistors and capacitors between the anode and the control grid cireuit of V3, constitutes a voltage negative feedback tone control arrangement, and is brought into modified action by means of the tone control switch Previously mentioned. Four positions-" Fidelity," "Brilliant," "Mellow 1," and "Mellow 2"-are dvailable on radio, and two positions-" Fidelity"


Figs. 2 (above) and 3 (below)-Top and botrom views of chassis, showing positions of various parts.
and "Mellow"-are available on gram., the seventh position actuates the receiver's main on/off switch.

These facilities enable the A.F. response of the receiver to be modified, and in practice the results achieved are quite effective. On the "gram" positions, however, a change takes place in the I.F. amplifier stage $V 2$.

## A Reffex Arrangement

Switches S14, S15 and S17 open, and S16 and S18 close. This produces an A.F. coupling from V2, in virtue of the resistive load R5 in its anode, through C7 and C8 and the volume control, to the grid of V3. The pick-up signal, being applied across the correcting network R6, R7 and C9 between chassis and the lower end of L14, thus finds its way to the signal grid of V2. This valve, therefore, acts as a voltage amplifier and raises the pick-up signal to a level suitable for fully driving V3.

## Other Features

Facilities are provided for using an extension loudspeaker. Switch $\$ 25$ brings the extension speaker into circuit, and switch $\$ 26$ mutes the internal speaker.

A double - wound transformer is employed in the power circuit, and thus the receiver can only be used on A.C. mains supplies.

Mains smoothing is catered for by the filter resistor R8 and the elctrolytic capacitors Cl and C 2 (a single-unit capacitor). In the radiogram version, R8 is replaced by a conventional smoothing choke; this is to minimise the residual hum which may otherwise become disturbing owing to the larger speaker baffle area of the radiogram cabinet.
Owing to the resulting rise in H.T. line potential, an additional resistive capacitive filter section is also included in the radiogram chassis.
The difference between' the tuning drive cord arrangements of the radio and radiogram versions is illustrated in Fig. 4.

## General Faults -

If excessive hum is experienced when the receiver is switched to the gram position, a $4.7 \mathrm{~K}!$ I-watt resistor should be added between the H.T. line and the top of R'S, and a $2 \mu \mathrm{~F} 350$-volt electrolytic capacitor connected from the chassis to the junction of the resistors. Later model receivers in this series embody this modification.
Insufficient volume when using a-high-impedance màg-
netic pick-up on the radio version may be due to excessive damping as the result of the correcting network, R6 and C9. Enhanced performance may sometimes be achieved by removing these components from across the pick-up terminals.

A general lack of sensitivity has frequently been proved, the result ol an alteration in the value of one


Fig. 4.-Method of fitting luning drive table models : (b) on the radiogram chassis.
or more of the 70 pF fixed tuning capacitors across the I.F. transformer windings. These capacitors are contained within the I.F. screening cans, and if poor sensitivity seems to be an elusive feature of the receiver, one is nearly always rewarded by replacing these parts. We would mention, of course, that the faulty capacitor can be located by adjusting the I.F. transformer cores in turn. If it is noticed that a considerable rise in sensitivity occurs when one particular core is adjusted, and that the peaking point is outside the range of the adjustment, then the associated capacitor should be replaced.

Weak signals, both on radio and gram, should lead one to check the condition of the $47 \mathrm{~K} \Omega$ resistor which is situated in the top-cap connector of the EBL31 valve. Since this resistor is subjected to considerable heat from the valve, it has a habit of either going very high in value or completely opencircuit.

If an intermittent "rumbling" noise becomes superimposed on the reproduction, the top-cap of the EBL31 valve should be scrutinised, preferably through a magnifying glass, for it often happens that oxidisation occurs between the grid wire leading out of the valve and the solder blob on the top-cap. This condition is frequently visible as a small black dot in the centre of the solder. It is easily cleared by quickly melting the solder on the cap, and while it is still liquid giving the valve a vigorous shake. This will remove the solder and permit the wire to be thoroughly cleaned, tinned and re-soldered-and in nearly all cases, save the expense of a new valve.
Apart from the $25 \mu \mathrm{~F}$ electrolytic capacitor in the cathode circuit of the EBL31 valve becoming shortcircuited, distortion on gram. rarely occurs. If it happens on radio, however, it is generally caused by C6 becoming leaky.
If the radio section fails, but the gram. section remains live (on the radio version this can be proved by switching to gram. and touching the live pick-up terminal with a finger; if a fairly loud hum is produced, then it can be presumed that the gram. section is operational), VI should, of course, be tested first, but if the valve and its associated electrode voltages check normal, and a click from the loudspeaker can be obtained by touching the grid ton-cap of VI with
a screwdriver, then in nearly all cases it will be discovered that the oscillator grid-coupling capacitor C10 is open-circuited.

Apart from becoming completely open-circuited, this capacitor sometimes increases in value. This has the effect of preventing oscillation on the long waveband, the medium and short waves working normally, thougT probably slightly of tume on the scale.

An intermittent crackling or "rustling" noise should lead one to suspect V1 and V2 for loose electrodes. These can be tested for this fault in sill by removing the aerial. tuning to a quiet part of the lons or medium waveband, and then gently but sharply tapping the valves in turn with the end of a screwdriver handle. If the valve has a loose electrode fault it will be revealed by a crackling or rustling from the loudspeaker as it is tapped.

When making a test of this kind, care should be taken to ensure that the noise is not being caused by poor contacts between the valve pins and the valveholder sockets; as opposed to tapping the valve, this can generally be proved by wriggling the valve in its holder.
Another source of valve noise is a poor connection between the metallised red coating on the ECH35 and the EF39 and the wire which connects this coating to the cathode pin. This generally results if the envelope of the valve is loose in its base. Noise from this source can be cleared up by binding about six turns of tinned 20 s s.w.g. wire around the valve between the envelope and the base. This will improve the electrical connection between the metallising and tine metallising connecting wire. The wire can be held firmly in position cither by soldering it at the ends, or by tightly binding over the top of it with insulating tape

## Alignment Procedure

The I.F. stages must be aligned first. To do this the receiver should be switched to the medium-wave position and the tuning set to the low-frequency end of the band (with the gang fully enmeshed). The A.V.C. should be made inactive by shorting C 6 , and the oscillator should be muted by shorting C4.

The volume control should be set at maximum, and a $465 \mathrm{kc} / \mathrm{s}$ signal applied to the top-cap of VI (the lop-cap connector should be removed and in its place a $\frac{1}{2}$ megohm resistor should be connected between the top-cap and chassis-this is necessary to maintain bias on the value). It is best to apply the signal through an $0.1 \mu \mathrm{~F}$ isolating capacitor, though the "earth". side of the signal generator can be connected direct to chassis (sec "Using Test Instruments "-which will appear later in the series). The signal generator should be modulated, and an output meter connected across the secondary of T .
The cores of L13, L14, L15 and L16 should be adjusted for maximum output (see Figs. 2 and 3). This concludes the I.F. alignment, and now the short circuit should be removed from C 4 , the resistor removed from VI grid, and the top-cap replaced.
The medium-waye section should next come under attention by applying a $1,500 \mathrm{kc} / \mathrm{s}$ signal, through a dummy aerial, between the aerial and earth sockets, and setting the tuning to correspond to this frequency ( 200 metres on the receiver scale). Before commencing operations, it is as well to ensure that the scale pointer is traversing the scale correctly, and that this
(Concluded on page 366.)

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# New Valve Construction 

DETAILS OF .SOME NEW MULLARD VALVES

THE well-known : advantages of screen-grid valves for high-frequency amplification has led to the ividespread use of tetrode valves in radio transmitters, In order to preserve the good performance of tetrodes at very high and ultrà high frequencies, special types of construction are necessary. Some recent Mullard double tetrodes, incorporating such features as internal neutra'lising, operate efficiently at frequencies as high as $600 \mathrm{Mc} / \mathrm{s}$.

The design of transmitting valves for use at freqưencies above $150 \mathrm{Mc} / \mathrm{s}$ presents a number of problems. Stray capacitances and inductances inside the yalve envelope affect operation more and more as the working frequency is raised. Despite these difficulties, a special construction now used in Mullard double beam tetrodes makes possible efficient operation at frequencies up to $600 \mathrm{Mc} / \mathrm{s}$.

For high-frequency operation, the screen grid valve has the very important advantage over the triode that its anode-grid capacitance can be made very small. In high-frequency transmitters, tetrodes and pentodes may, therefore, be operated in conven-tional-circuits without neutralisation. As the working frequency is increased, however, the effect of stray inductance in the screen and cathode circuits becomes marked. Inductance in series with the cathode gives rise to degenerative feedback which results in a lowering of the valve input impedance. This causes a waste of drive power. Inductance in series with the screen grid can give rise to positive feedback which may result in instability.


Figs. 1, 2 and 5.-Cn and Cn ${ }^{1}$ in Fig. 5 are neutralising capacitors.


Fig. 6.-Two of the new valves.

## An Early Solution

The earliest remedy to this problem was to incorporate two screen-grid valves in one envelope, with the screens and cathodes connected together by lowinductance straps, the centre point of a strap being brought out as a piñ connection. A typical circuit arrangement, showing stray inductance is illustrated in Fig. 1.

The existence of two separate electrode structures side-by-side in the envelope necessitated rather long grid and cathode straps, however, and these possessed sufficient self-inductance to cause undesirable feedback at still higher frequencies. At these higher frequencies, Fig. 1 ceases to be an adequate representation of the circuit, which becomes more like Fig. 2 . The effects of $L k, L k^{1}$, and $L s, L s^{i}$ are not cancelled by the push-pull connection.

In the current range of Mullard V.H.F. double tetrodes an improved method of construction is used to reduce lead inductance to the absolute minimum, and thus permit operation at frequencies as high as $600 \mathrm{Mc} / \mathrm{s}$. This construction is illustrated in Figs. 3 and 4. It will be seen from Fig. 3 that a single indirectly-heated cathode is employed. This is of roughly rectangular shape, and only the long sides which face the grids are coated with emissive material. In effect, there are two separate cathodes, interconnected by the short sides of the rectangle, which act as very low inductance straps. A single screen grid is placed round both grid-cathode systems, completely climinating the effects of screen-lead
self-inductance. In practice no screen decoupling capacitor is needed and the screen may be connected to its high-tension supply via a choke or resistor.

Since this type of construction virtually eliminates the effects of screen and cathode lead inductance, the only remaining cause of instability at V.H.F. is the small residual grid-anode capacitançe. It is possible for an amplifier to become unstable at certain high frequencies as a result of feedback through this capacitance. The effect is easily eliminated in a pushpull stage by connecting neutralising capacitors from the anode of each valve to the grid of the other. If, however, these capacitors are connected externally, the presence of stray inductance in anode and grid leads and the leads of the capacitors themselves, has the effect of upsetting neutralisation at high frequencies. This is illustrated in Fig. 5, in which the stray inductances are indicated.

## Internal Condensers

In Mullard V.H.F. double tetrodes, the effect of the grid and anode lead inductances in respect of neutralisation has been eliminated by incorporating neutralising capacitors inside the valve. These take the form of small pieces of wire, indicated in Fig. 4 as C and C $^{1}$, connected to each grid support, each extending to a position near the anode of the other


Fig. 3.-Sectional vien of the electrode assembly.
tetrode. The result is a true direct electrical connection of electrode to capacitor without intervening stray inductance, and neutralising is effective at all frequencies.
The QQVO6-40A (CV2797) is a larger valve, with
an anode dissipation of 20 watts per anode-twice that of the QQVO3-20A. Under telegraphy conditions it gives at full ratings a load output of 72 watts at $200 \mathrm{Mc} / \mathrm{s}$. With reduced ralings, outputs of the order of 45 watts at $500 \mathrm{Mc} / \mathrm{s}$ are obtained.


Fig. 4.-The actual valve which is shown in sectional form in fig. 3.
These valves have certain interesting construttional features apart from the special electrode configuration described above.
The anodes are of molybdenum coated with powdered zirconium, which reduces secondary emission, improves radiation of heat, and acts as a getter, with the important advantage that its ability to absorb gases increases as the temperature is raised. The anode lead-out wires are made thick in order to reduce the adverse effects of self-inductance mentioned above.

## SERVICING RADIO RECEIVERS

## (Continued from page 362)

is properly synchronised to the movement of the vanes of the tuning gang; the tuning mechanism should be adjusted if necessary.

The medium-wave oscillator trimmer T4 (Fig. 3) and the medium-wave aerial trimmer T 2 (Fig. 2) should be adjusted for maxinum output. The calibration should be checked at $600 \mathrm{kc} / \mathrm{s}$ ( 500 metres). which, provided C5 is up to standard, should be found to be within a reasonable tolerance.

The long-wave oscillator trimmer T5 should next be brought into adjustment by applying a $250 \mathrm{kc} / \mathrm{s}$ ( 1,200 metres) signal, and setting the receiver accordingly. The trimmer should be adjusted for maximum output. Correct tracking should be
maintained over the entire band, provided the fixed padder $C_{5}$ is of correct value.

The short-wave band should next be aligned by adjusting the short-wave oscillator trimmer T3 (Fig. 3), and the aerial trimmer Tl (Fig. 2) for maximum output at $17.14 \mathrm{Mc} / \mathrm{s}$ ( 17.5 metres).

Tracking should be checked at $6.98 \mathrm{Mc} / \mathrm{s}$ (43 metres), and if a considerable crror exists the turns spacing of the short-wave oscillator coil L10 should be altered slightly as a method of compensation. In order to achieve optimum. sensitivity over the whole of the short-wave band. the turns spacing of L4 should also be adjusted. After making coil adjustments of this kind, it will be necessary to repeat the short-wave alignment process until no further improvement is registered.

Finally, the shorl-circuit should be removed from the A.V.C. line.


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# Surplus Vibrator Power Units 

DETAILS OF SURPLUS UNITS WHICH ARE AVAILABLE

By E. G. Bulley

THERE are available upon the surplus market various types of these units, some of which have been removed from other equipment, such as mobile transmitters and receivers, whilst others are new and unused and, in fact, still in sealed cartons.
Such units are extremely useful for supplying the power necessary for car radios or other similar mobile electrical equipment. The input power source for these units is mainly 6 and 12 volts. One will, therefore, appreciate that such voltages can be furnished by a car battery.

Fig. 1. - Circuit of a non-synchronous vibrator unit.


Now, at the same time one will appreciate that as the reed has made electrical contact with point "A" it has also short-circuited the energising coil, and thus the coil loses its magnetic property. The sprung reed is now released and carried by its own inertia to make contact with point " B." With the reed in contact with " $B$," the current now fows through the other half of the transformer primary. This cycle continually repeats itself, ". the reason being that as contact is broken at point "A" the shortcircuiting effect of the coil is cut out and results in , the coil once again becoming energised.

This interrupting phenomena

There are, however, two types of these units available, one incorporating what is known as a synchronous vibrator, and the other using a nonsynchronous unit. Typical units of the former consist of a vibrator, specially designed transformer, various capacitors and R.F. chokes: and this type of unit is self-rectifying. The other type of unit, namely the non-synchronous, employs a valve or metal rectifier in addition to the components already mentioned for the synchronous unit. The additional parts are required as being that the vibrator is of the non-rectifying type, as will be explained in the following paragraphs.

## Non-synchronous

The non-synchronous vibrator consists of a single reed set equidistant between two contacts. The arm upon which this reed is assembled is secured at one end in a metal frame, and the other end is in the vicinity of a magnetic coil. To assist the reader to understand how a tibrator operates, reference should now be made to Fig. 1.

When the switch (SWI) is made, a specified voltage is applied to the coil of the vibrator, the coil becomes energised and a magnetic field is created. Under the influence of this the arm on which the reed is assembled is attracted. The reed thus makes contact with contactpoint "A," and by so doing the current flows through half the primary transformer winding.
wave gas-niled rectiner having what is termed an ionic heated cathode, and does not therefore require a heater voltage. Such units are extremely useful for car radios. The other types are conventional vacuum rectifiers and need no further comment.

## Synchronous

In the synchronous units similar components are used, with the exception that a synchronous vibrator is employed. This eliminates the necessity for a separate rectifier. A typical circuit of such a unit is shown in Fig. 2. The reader will, therefore, by making reference to it, appreciate the fact that the vibrator here has an extra pair of contacts. These contacts are connected to the secondary winding of the transformer so that when the voltage is induced into this winding it is likewise fed back and rectified by the vibrator, the positive D.C. output being taken


Fig. 2. - Circuit of a synchronous vibrator.
from' the centre tap of the secondary winding.
Such units, however, in common with the nonsynchronous types, use R.F. chokes and capacitors. RFC1 and C1, RFC2 and C2 are filters, their presence being to prevent R.F. being passed to the circuit in which the unit is eventually connected. C3 is, however,


It may be found, however, that the rating and value has been obliterated, and in that case the correct value of the condenser can be determined by trial and error with an oscilloscope.

This is done by connecting the vertical plates of the oscilloscope across the primary winding and making a study of the waveforms. Typical waveforms are shown in Fig. 3. The reader must bear in mind that too large a buffer condenser will result in excessive wear on the vibrator contacts, whereas too low a value will cause arcing to occur at the faces of the contacts and thus reduce the life of the vibrator. As a matter of interest some units available
an important component in the unit and is essential to the correct operation of the unit. This condenser is termed the " buffer," its purpose being to provide the correct time constant for the circuit, and absorb any surges that may occur when the vibrator contacts make and break.

The value of this condenser is very critical and one should, therefore, bear in mind that if it is intended to replace this condenser, should it be faulty, only one of the correct value and rating should be employed.
upon the surplus market have the buffer condenser connected across the primary winding: this, however, is usually. only found in those utilising 32 -volt vibrators.

Condensers C4 and C5 form parfor the smoothing filter which is conventional, but it is as well to mention that a great number of the units available do not have their own smoothing : one must, therefore, add these to such units before incorporating them in any receiver, etc.

## A Simple Neon Tuning Indicator

THOUGH magic eye and meter-tuning indicators are excellent and largely used, the cost of the "eye" or meter may prevent some constructors using these devices. Where a simple tuning indicator is required, at very low cost indeed, a neon bulb may be used in the circuit shown in Fig. 1. These neon bulbs may be obtained new from Osram and other makers, or purchased at extremely low cost from some ex-Service stockists. In the latter case, the ex-Service $10 \mathrm{E} / 327$ is suitable.

The component values shown were most suitable for a receiver with 250 volt H.T. line and 6 K 7 inter-mediate-frequency stage. Best results of all will be obtained by using two pre-set potentiometers or resistors of about 15 K and 500 K maximum value, so that circuit constants can be adjusted for the desired effect. The 5 K resistor (or 15 K pre-set resistance) is wired in the H.T. line to the I.F. stage, the .01 to $.1 \mu \mathrm{~F}$ condenser being for by-pass purposes.

The type of neon shown would not operate if connections to it were reversed. As it requires a large Edison screwholder, difficult to obtain, the glass bulb was made a push-fit in the panel, and leads soldered directly to the bulb as indicated.

As a station is tuned in, the voltage drop across the 5 K resistor falls, due to decreased valve anode current. The voltage applied to the neon thus increases. The receiver is therefore tuned for maximun glow. If pre-set resistances are used, the neon can be set so that it just fails to strike when tuned between stations. It will then begin to glow when a station is approached, and be brightest at the point of correct tuning. It will not operate with weak stations. A correct tuning indication can, however, be obtained on all the most powerful stations.

The circuit cannot be used with small sets having a low H.T. voltage. For receivers with higher H.T. voltages, or using different valves, the adjustable circuit in Fig. 2 may be employed. The .5 megohm potentiometer slider should be turned towards H.T. negative until the neon strikes. The degree of brightness change on tuning can then be adjusted by the 15 K resistor.


Fig 1 (left) --Showing the neon tuning inticator unit. Fig. 2 (right).--An adiustable circuit.


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#  <br> A DX AID <br> DETAILS OF A SIMPLE AUDIO FREQUENCY LIMITER AND FILTER WHICH CAN BE ADDED TO AN EXISTING RECEIVER TO IMPROVE R.T. RECEPTION, PARTICULARLY ON HEADPHONES 

As the reader will probably wish to incorporate this circuit in existing apparatus rather than build a separate unit, no constructional details are shown. There are, however, no snags in the wiring, though the A.F. choke should be positioned to avoid any possible hum induction.

M
OST amateurs will agree that headphones are essential to pick out those elusive DX signals which even the "family QRM" would drown if received on a loudspeaker.

For the purpose intended-for use with head-phones-the type to be described was found to be really efficient, simple and inexpensive to build, and having the advantage that it can be fitted to any type of receiver.
Basically it consists of a high-pass filter, followed by a full-wave clipper (set manually), and finally a low-pass filter.
The purist will probably argue that this circuit will introduce considerable distortion, but in practice even with maximum clipping (about 20 db )-when the output monitored on an oscilloscope would consist of square waves !-speech remains crisp and intelligible.

Bass response is attenuated before the clipper (by choice of coupling components) but maintained afterwards in order to preserve gain and reasonable quality. The low-pass filter used after the clipper is necessary to remove objectionable harmonics, but is also useful to restrict the audio frequency range to that required for speech.
In practice the "clipper level control". is set so that audio distortion is just noticeable with normal volume setting.
Another useful feature of this circuit, which will be more than appreciated by users of straight sets without AVC, is that the
audio output cannot exceed the sets without AVC, is that the
audio output cannot exceed the

To others, headphones are essential for a different reason-the music of an occasional "VK" at " R9+" is classified by the rest of the family under the general heading of " noise."
What all headphone listeners share in common, however, are repeated attacks of "tortured eardrums." When straining to hear a weak signal with the controls "all out " even a light switch operated in the vicinity can cause a click sufficient to give actual pain.
An efficient " noise or crash limiter " is clearly the answer and a filter to remove what is not wanted from the received signal. -

## The Circuit

Several types of noise limiter were tested, from a simple "saturated pentode" to a "compound fullwave " type operating automatically on the carrierwive levela.

## LIST OF PARTS

VR1-5K Wirewound Pot.
VI-6H6, ER91, etc. Any D. Diode valve will suit. Xtal diodes not R3-1 meg $\Omega$ watt carbon. recommended. C5-.002 $\mu \mathrm{F}$ paper Tub. C $6-50 \mu \mathrm{~F} 12$ volt Electro lytic.
$\mathrm{Cl}-.001 /$ /F paper Tub. $\mathrm{C} 2-.002 \mu \mathrm{~F}$ paper Tub. current). current). R1-100K $Q \frac{1}{4}$ watt carbon. R2-100k $\Omega$ watt carbon $R 2-100 k \Omega$
$R 3-1$ meg $\Omega$ watt carbon. suit. Xtal R4-l meg $\Omega$ i watt carbon. $V 2-6 J 5$-or output vaive pre-set level: so when tuning through the band from a weak station to a powerful one, the agonising moments before the volume control is re-adjusted are spared.


Graph showing response to be aimed at per $R / T$ reception.


Graph showing how audio naveform is clipped on high-anplitude signal or noise pulse.

# Programme Poillerers 

Moderate Askey

TרHE Arthur Askey shosw, " Hello Playnates," struck me as very rioderate. Composed to precisely the same formula as all the others of its kind, "Big Hearted Arthur" did not seem half as funny as when he has seven or eight minutes all to himself, or very nearly so. The supporting cast of David Nixon, Diana Decker, Irene Handl and Pat Coombs, with Bob Sharples and his music, were about the same-fair to average. This type of show must have played itself out more or less; that is, pending the arrival of another IIma and àll who contributed to it. All we can expect is current events and personalities gagged with a bit of music thrown in to make two breaks. At least, that is all we get. Two of the "jokes" in " Hello Playmates" the last time I listened to it will serve to show the general level of its humour. "The only dog food that tastes like a postman's leg," and "When she sings 'I love Paris,' you can see the postcards in her eyes." Hardly another lima!

## Saturday Jazz

The Saturday afternoon jazz programmes at 50 'clock make quite a remarkable series.
Goodness knows how long they have been going on now. I know of no other branch of music which has been treated over such a length of time so exhaustively or, if I may so, with such skill. They are very interesting and serve to show up one remarkable fact very plainly. The old records of dance bands of, say twenty years ago, are much more playable and "easy on the ear" than are records of "classical" music played on orthodox instruments and by orthodox combinations. Some of these latter sound frightfully primitive these days and that is, presumably, a sufficiently good reason for not playing them to any extent. Dance band recording seems to have been very superior to straight music in pre-electric days.

Of the many brilliant personalities whose work is featured in BBC programmes from time to time, few can rival Peter Scott for erudition, entertainment and microphonic efficiency. Son of the immortal Antarctic explorer, Mr. Scott is a naturalist of eminence and wide experience, and whether telling the children in Children's Hour about a bird's nest in Lapland, or their elders of his work in the bird sanctuary on, I think, the Severn, he is always one of the most welcome of broadcasters. He is that rare person, an expert at his job-and what a fascinating one-and a charming man.

## Plays

One of the most rubbishy and unworthy plays ever offered on a Monday evening was sandwiched between two masterpieces. This was unfortunate in itself as it only served to show up its many weaknesses even more than if broadcast on its own.

I refer to "Christopher's Day," an adaptation by James McFarlan of Edwin O'Connor's story "The Oracle." The story, of a vanity-stuffed news commentator, who scales tawdry heights and ends in

## Our Crilic, Maurice

Reeve, Reviews Some

## Recent Programmes

murky depths, was of the feeblest description. Throughout the 90 minutes of its wearisome journey, we were plagued with the "Chris baby" of the floozic he was foolish enough to take up with. Any man with brains would never have looked in such a creature's direction. Or, having regrettably done so, would have been forced to shoot either her, or himself, or both. The piece was set in America though the assumed accents were far from convincing. James McKechnie, Tucker McGuire, Aletha Orr and many others learnt themselves to it all.

The two masterpieces were, of course, Shaw's "Major Barbara " and lbsen"s " The Master Builder." Would anyone have much chance with two such escorts? Irene Worth, Barbara Couper, Anthony Jacobs, Frank Pettingell, Hugh Manning and others gave themselves to Shaw, and Malcolm Keen, Gladys Young, Cyril Chaps, Preston Lockwood, George Merritt and Ursula Howells to Ibsen.
" Major Barbara" was in The Stars in Their Choices series. Good drama merits, and usually receives, a worthy performance as I have often pointed out in this column. There was some excellent acting in both these productions, notably by Mr . Pettingelt in Shaw and Mr. Keen in Ibsen.
"Caroline" was another Maugham tribute. It is surprising how many of the master's early works date, unlike Wilde, who has become a period classic along with Sheridan and Goldsmith. Perhaps we are still too near the times and the fashions. But the epigrams roll out and never fail to amuse and the theatrical craftsmanship is always fascinating. Roger Livesey and Ursula Jeans played it with unfailing charm and finesse. Richard Hurdnall, Noel Hood, Leslie Phillips, Avice Landone and Mairhi Russell adorned the cast. How like Jimmy Edwards Mr. Livesey frequently sounded.
" The Dover Road," by A. A. Milne, was ano her delightful play in The Stars in Their Choices series. Telling of an eccentric and wealthy bachelor living in great comfort hard by the road, who stops cloping couples and, on one pretext and another, invites them in to play host until he has convinced them of the "error" of their intentions. Many will remember Henry Ainley in the part at the Haymarket. It was extremely well acted by Laidman Browne, assisted by Brian Haines, Joy Rogers, John Gabriel, Lucille Lisle and Richard Bebb.

## American Pianist

Julius Katchin is a brilliant young pianist from the United States who is deservedly making a name for himself. He did full justice to the fabulously difficult Brahms in D minor.

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> EFSJ $\begin{aligned} & 6 \mathrm{~V} \\ & 6 \mathrm{~J} 5\end{aligned}$ 6S.J7 $\begin{aligned} & \text { 6K8 } \\ & \text { VU134 }\end{aligned}$ ${ }_{8}^{807}{ }_{5}$ EF37 $\begin{aligned} & \text { 5Z4 } \\ & \text { EL50 }\end{aligned}$ $\begin{aligned} & \text { EL50 } \\ & \text { CV183 }\end{aligned}$ CV287 VR116 | $8 / 6$ | $6 B W 6$ | $8 / 6$ | $5 R 49$ |
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## R1155

SIR,--Until recently I have been very pleased with the performance of my R1155. Then one day I noticed that there was considerable distortion on deeply modulated signals. Once noticed, this has become a source of continual annoyance. I wonder if any other readers have noticed this on their 1155? I have tried all the usual remedies for possible causes : careful check of all valve voltageswithin the scope of my home-made voltmeter, 2,000 ohms per volt-and check of the filtering on the A.V.C. lines and of the alignment. But nothing seems to make any difference. The symptoms could be A.V.C. distortion or intermodulation; but like Jerome K. Jerome, each time I read of an ailment new to me, I feel sure that is what I have got.
$I$ feed a tape recorder from the top end of the A.F. volume control through a 0.1 condenser and 1 meg. resistor in series, and with a 100 K resistor in parallel with the output socket to H.T. negative. The output socket to the recorder is fitted in a spare hole in the top left hand corner of the receiver panel, vacated by the meter balance control of the liss. This arrangement works very well and gives an output independent of the receiver volume control when using A.V.C. though not on " manual" when R.F. volume control is used.-J. P. Marchant (Bedford).

## $\therefore \quad$ Amplifier Design

$S^{I R},-I$ read with much interest and more than a little surprise Mr. Kerslake's excursions in the field of audio-engineering, and in view of the-to say the least-astounding "practical" results he has achieved, I am surprised that he should wish Mr. Hindle to "evolve" a circuit.

I myself have been trying for years to achieve such realism, with theorists like Messrs. Williamson, Baxandall, etc., for guidance, but I see now that my efforts should have been empirical, and that theory must always take second place to good aural results.

In conclusion, therefore, I would like to invite Mr. Kerslake to present to an excited and expectant audio-world the exact circuitry, operating voltages and apparatus necessary for us to achicve all these "brilliances " and "realism in music."
Since Mr. Hindle has been invited to "deal" with the cathode-follower output stage, perhaps he also would care to comment on the above, possibly including what Mr. Kerslake calls the Hotse and Pony output.-W. A. Cornish (Leicester).

## Unpopular Valves

SIR,-With reference to your " economical quality receiver," the saving in components, space and cost in using the double-diode-pentode valve 6 B 8 as

amp. is very worthwhile. Quite apart from doing away with two sets of bias resistors and condensers (one an electrolytic), the lowering of H.T. and L.T. current consumed, makes the use of physically smaller mains transformers possible, so that chassis and cabinet sizes can be reduced.

Consider the saving in cost: a 6K7 I.F. amp. stage plus 6Q7 needs :

| 2 valves | say $15 /-$ |
| :--- | :--- |
| $2 \mathrm{v} /$ holders | say $1 /-$ |
| 7 resistors | say $2 / 4$ |
| $2 \times .1 \mu \mathrm{~F}$ condensers say $1 /-$ |  |
| $1 \times 25 \mu \mathrm{~F}$ condensers say $2 / 6$ |  |

$$
21 / 10
$$

A count of the cost in the third stage of the circuit referred to is most revealing, especially as two advertisers offer 6B8, in the April issue, at $4 /$-each !
This valve can also be used as a direct substitute for 6Q7 or EBC33, provided that the anode and screen are strapped together.

Another cheap space saver is the little 9003, which I always now use instead of 6 K 7 , wherever its. 15 amp . heater consumption permits.-Richard Page (B.A.O.R.5).

## A Peculiar Fault

SIR,-I recently had an experience I feel would be of interest to participators of Practical Wirfless especially. The experience was concerned with the Ferguson model 208V. After the set had been switched on for about half a minute, and as the valve-heaters were warming up, the pilot lamp, which is in series with the valve-heaters, would gradually light up and then would "blow." This state of affairs pointed to a heater-to-cathode leakage in the rectifier valve being great enough to step up the heater-pilot-lamp, voltage, and thus cause the pilot lamp to "blow." The valve in question was the UY41 Mullard half-wave rectifier. A new UY4I Mullard valve was substituted and all was well. The original UY41 valve was subsequently , tested and found that a cathode-to-heater "short" developed as the valve was in the state of conducting.-J. H. Robinson (Thornton, Fife).

## A.C.|D.C. Apparalus

SIR,-Quite often you publish some excellent " live chassis" circuits in your magazine. There are many newcomers to the fascinating hobby of radio construction quite unaware of the dangers involved when using A.C./D.C. receivers, etc.

Perhaps you could publish some basic rules regarding these sets, i.e., no metal parts to be accessible to the user, knob screw holes to be plugged. Chassis securing bolts beneath the cabinet are often overlooked and very few circuits include fuses in the
mains input. Mains droppers in confined spaces when using a wooden cabine! are a fire risk, too.
As a service engineer I often mect frightening " hook-ups," such as extension speaker connections to A.C./D.C. sets, no back panels, poor flex on 15 amp . plugs and stone floors to "earih" oneself. -A. E. J. Simons (York).

## An Obscure Fault

SIR,-I have a receiver built from the Osmor " Q " coil pack type H.O. diagram, the valve line up being $6 \mathrm{~K} 8,6 \mathrm{~K} 7,6 \mathrm{Q} 7,6 \mathrm{~V} 6,5 \mathrm{Z4G}$.
The set suddenly stopped working, and as I had a complete change of working valves from another set I replaced valve for valve.

Fault (or, rather, unpardonable error) No. 1 was that one set of valves included a metalised 6 K 7 and the other did not. Somehow 1 managed to replace the 6 K 7 with a 6 Q 7 and vice versa.
Having corrected that fault 1 discovered that the heater of the $6 \mathrm{K7} 7$ (from the original set of valves) had burned out, but iss replacement only gave me a slow pop-pop noise which I believe is called motorboating.
Touching the grid of the 6K7 stopped the noise and brought in the station, though somewhat weakly
It was some time before it dawned on me just why the signal was not getling through.
The new 6K7 was of the metal can variety, and the earth pin had not been connected whilst originally wiring for such a replacement. The fault was immediately cured by earthing the pin which connects the casing to the valve holder, thereby removing the stray capacity.
As I said before, this is not an obscure fault, but it might be of use to Mr. Apps when he writes another of his interesting articles.-J. Crichton Bell (Southsea).

## A Satisfied Reader

SIR,-I have for some time intended writing this letter, and the current issue (May) of Practical Wireless has finally prompted me to do so.

The very wide range of subjects covered in your publication, from radio-control for models to the construction of a gram, is an endess source of pleasure. One can pick up any issue of Practical Wireless and be pretty certain that it will contain an article on at least one radio subject in which one is particularly interested.

The articles are always well written in a style which permits of easy assimilation by the non-tectnical, at the same time containing sufficient facts and information to satisfy the more advanced reader.

I particularly like the way in which circuit diagrams are laid out and drawn ; your artist in this respect is, in my opinion, streets alhead of those of your conlemporaries.

There is one small detail 1 should like to bring to your attention.
In a recent issue of Practical Wireless there appeared an article on Pi-Network TX output
circuits by O. J. Russell. B.Sc. The same article appeared in at least two other radio periodicals though in a slightly altered form, of course.-J. D. Pearson (Barrow-on-Humber).

## Modifying the R1132A

$S^{I R}$,-May I reply to the letter of Mr. David W. Button, of Bedford, in the May issue of Practical Wireless.
Mr. Button suggests that interfering with the coils of the RII32A is a very tricky job. I do not agree with this statement at all. I merely renoved the existing coils, and wound new ones and soldered these in position. True, there was the usual trial and error of squeezing up, and opening up of the turns to get the best results, and trimming to put the required frequencies on the dial where I wanted them.
A tiered circuin does not work as efficiently at either minimum or maximum, as it does in the centre of its range: therefore the most used frequencies should be trimmed to the centre or near the centre for best results. A 1132A Rx can be modified to an 1481 or 1526 by merely altering the coils. Providing the $R x$ is working before the modifications are undertaken, there is no trouble at all.

1 have seen 1132A converted to 1481 and 1526 with coils wound with 22 s.w.g. copper to 18 s.w.g. iroil wire and work all right. I am using one at the moment with $13 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. phosfor bronze wire, which is very hard and difficult to use for winding coils of fin. I. diameter, nevertheless it is working perfectly.

I see no reason why the trimmers should be changed at all unless they have been damaged. I have modified six of these $R x$ which are all working correctly as far as I know. Two people I know of have modified these sels with no trouble.

May I also add a few lines to J. F. B. Elder (Smalley) ie time and frequency checks. Why not tulle to MSF $2.5,5,10 \mathrm{Mc} / \mathrm{s}$, which gives vocal time checks every 14 minutes, also $1,000 \mathrm{c} / \mathrm{s}$ tone and one-second pulses? Clocks can be set at any fiveminute period throughout the 24 hours. Frequencies ale accurate to within $\pm 2$ parts in $10^{8}$ of the nominal values,-L. H. Cox (S.W.18).

## American Test Equipment

SIR.-Your correspondent signing "Grayshaw instruments" (Harpenden) (page 282, May. 1955, issue) entirely misses the point, which is that imported kits of parts are arailable to purchasers.
I have personally tried to buy test equipment bearing the trade name used by your correspondent, at a dealer who claimed to be the agent for the apparatus. This dealer had no slock to dispose of but did exhibit what purported to be a signal generator and which may or may not have contained innards.

A collcague of mine forwarded cash with an order for another widely advertised piece of test equipment which he received after seven months, having repeatedly written without reply in the interim and eventually being obliged to resort to legal advice.-M. L. Elliott (Fareham).


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