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## INDEXES

Indexes to Vols. 35 and 36 are available at 80 p each from the Editorial Office (address above)

## QUERIES

We regret that we cannot answer technical queries over the telephone nor supply service sheets. We will endeavour to assist readers who have queries relating to articles published in Television, but we cannot offer advice on modifications to our published designs nor comment on alternative ways of using them. Correspondents should enclose a stamped addressed envelope.
Requests for advice on dealing with servicing problems should be directed to our Queries Service. For details see our regular feature "Service Bureau". Send to the address given above (see "correspondence").

## this month

333 Leader<br>334 Practical Computer Programming, Part 3<br>This time a look at computer languages, what they do and how they go about it, with some common programme writing routines.<br>336 More Troubles<br>Les Lawry-Johns<br>Never a dull moment - either the sets or their owners bring problems along.<br>338 Wide-range Capacitance Bridge<br>David Botto<br>Designed with the needs of the service department in<br>mind, this bridge has five ranges covering 5 pF to $2,000 \mu \mathrm{~F}$, with a sixth range for matching resistors, capacitors,<br>etc. Resistance ranges can easily be added. There are also<br>two squarewave test signal outputs. The unit is easy to build and uses readrly available components.<br>344 Letters<br>347 A Low-cost TVRO Installation, Part 1<br>Roger Bunney<br>The aim was to achieve an efficient satellite TV receiving system at minimal cost, sacrificing ease of operation to optimum performance. Details of the equipment and a method of adapting a patio mount to obtain azimuth adjustment.<br>350 TV Faułt Finding<br>Reports from Mick Dutton, D.H. Davies, Hugh<br>MacMullen, Joseph Cieszynski, Roger Burchett and Philip<br>Blundell, Eng. Tech. Plus a note on variac repair by John de Rivaz, 8.Sc. (Eng.)<br>eletopics<br>News, comment and developments.<br>354 Long-distance Television<br>Roger Bunney<br>Reports on DX conditions and rception plus news from<br>home and abroad.<br>356 A Professional Institution for TV Technicians<br>Ian Channing<br>The Society of Electronic and Radio Technicians'<br>Incorporated Practitioners in Radio and Electronics<br>Division provides a professional service for servicing personnel.<br>357 Fast-shutter Video Cameras<br>Eugene Trundle<br>Many video cameras with solid-state image sensors<br>now have electronic shutter operation to minimise blur<br>with fast-moving subjects. How the sensor and shutter system work.<br>\section*{The Art of Servicing}<br>B. A. Berry<br>The art lies in the diagnostic steps that can be taken to localise the cause of a fault before any test equipment is brought into play.<br>363 Next Month in Television<br>364 VCR Clinic<br>Reports from Steve Beeching, T. Eng., R.S. Narwan,<br>Khalied Kwimry, Christopher Holland, Alfred Damp and<br>Eugene Trundle.<br>Sound Systems, Part 3 Geoff Lewis, B.A., M.Sc.<br>Details of various digital sound systeris including<br>Dolby ADM, NICAM 728 and the MACIPacket variants.<br>370<br>Service Bureau<br>371 Test Case 303<br>OUR NEXT ISSUE DATED APRIL WILL<br>BE PUBLISHED ON MARCH 16



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## COVER PHOTO

This month's cover photograph shows the wide-range capacitance bridge in use on the bench - see article on pages 338-342.

## CORRECTION

We apologise to Peter Richards for an error that occurred in his letter in the January issue. Velocity is measured in metres $\mathrm{sec}^{-1}$ or metres/sec. Metres/ $\sec ^{-1}$ as printed is incorrect.


## The Japanese Onslaught

Yours truly does not subscribe to the determinist theory of history, preferring the cock-up theory - that accident and miscalculation play a large part in determining the course of events. This seems to be self-evident since even the best laid plans are prone to come unstuck. There have, nevertheless, been times when one has been tempted to think that the onslaught by Japanese consumer electronics manufacturers on the European market has been part of a deep-laid plan, possibly co-ordinated by the at times sinister looking hand of the Japanese Ministry of International Trade and Industry. Quite how well co-ordinated it has been we'll probably never know. We certainly know its effects, but when these are analysed you tend to get back to the general muck-up view of things.
A powerful article in a recent issue of the Financial Times, by Fred Burton of the University of Manchester Institute of Science and Technology, argues that Europe has in the main been following self-defeatist policies in the consumer electronics manufacturing field. He questions the advantages claimed for encouraging investment by Far Eastern manufacturers in European production facilities - that local production is enhanced, jobs are created, exports are increased and that benefits accrue from technology transfer and the effects Japanese manufacturers have on local component suppliers. He points out for example that employment in consumer electronics manufacturing in Europe fell from 250,000 in 1975 to 120,000 in 1985, adding the qualification that productivity increase contributed to this. He suggests that technology transfer is insignificant - that although Japanese companies account for 20 per cent of CTV and 90 per cent of VCR production in Europe (before Philips VHS machines?), research, design and development are all carried out in the Far East - fewer than fifty Europeans are engaged in development work for Japanese concerns, none on research. Technology transfer, he concludes, "is confined to job training and technical instruction to suppliers". He feels that encouraging Japanese manufacturers to set up in Europe by giving them incentives and subsidies has been of only short-term benefit, the main aim being to achieve import substitution.
When you look at what has happened in the UK one is inclined to be suspicious. Toshiba took over the old Rank manufacturing facility, Hitachi did the same with the GEC plant, Sanyo likewise with Pye and Mitsubishi with the Tandberg (UK) plant. Sony, Panasonic and others have started up afresh on green-field sites. But the fact is that of the plants that were taken over by and large only the buildings were retained, the old manufacturing facilities being stripped out. One could argue whether all this would have happened had the UK industry been fundamentally sound, with adequate investment, modern production facilities and, the end product, well designed equipment built to last. For various reasons that belong to history, the UK consumer electronics industry was not in a particularly healthy condition when the Japanese onslaught started. What's left of it is rapidly being taken over, again largely by overseas companies, in an attempt to salvage something - one thinks of Thomson's take-over of the Ferguson CTV plants and the current question mark over Fidelity (see Teletopics).
What the Japanese have all along sought is a presence in Europe to avoid possible trade barriers and duties - the same policy was followed in the USA, where the major Japanese manufacturers long ago set up manufacturing facilities. The concern of Japanese manufacturers at the possibility of being excluded from the European market is understandable. Have their efforts been successful? As an insurance policy maybe, but financially the answer seems to be no. To return to Fred Burton's research, he comments that "throughout europe Far Eastern subsidiaries have shown a return on sales below three per cent, with Sony, Hitachi, Sanyo and Mitsubishi declaring large losses on their UK operations". The time scale of this economic performance is not stated, but the fact remains that these European plants have hardly been a resounding success as regards profitability. Perhaps this is once again a part of the Japanese view that market share is the all-important thing.

Fred Burton concludes that to transform Far Eastern investment in Europe so that it contributes to the long-term economic welfare of the region various conditions should be laid down, for example that there should be a greater research and development element. It's difficult however to see how this could be put into effect.
Europe has not entirely lost out in the TV field. The development of the MAC satellite TV transmission standard and its implications for receiver manufacture, work on videotext systems, digital sound systems and tube technology (45AX) are all of major significance, comparable to anything being achieved in Japan where the emphasis at the moment seems to be on digital video processing (a field where Europe led initially). Now that European setmaking has been substantially rationalised - the take-overs by Thomson and Nokia and Philips' collaboration with Grundig - the manufacturing side of the industry should be a lot healthier. There doesn't at present seem to be too much cause for alarm. Basically what I'd say we've seen over the years has been a messy series of moves by various companies trying for either short- or long-term success, in other words the muck-up theory of events. What could, in retrospect, be said is that had a more determined effort been made to rationalise UK TV setmaking back in 1979, when the National Economic Development Council produced a plan, we might still have had an indigenous UK TV industry. But then again we might not: remember the grossly overvalued pound in the early 80 s?

# Practical Computer Programming 

## Part 3

Mike Phelan

Having discussed microprocessors and operating systems without going into fine detail it's time for us to tackle the subject of computer languages. A microprocessor chip feeds itself with a stream of highs and lows, i.e. ones and zeros, which it finds in various memory locations. It's told where to look by these highs and lows, the process being started by a program held in a permanent piece of memory (ROM). The first instruction will be found at a fixed location (address) that's used by the particular type of microprocessor. The computer's operating system looks after directing information to the screen, printer, disk or tape, etc., and running the various programs.

The computer language BASIC, which most home computers use in various forms, is itself held in ROM and is loaded when the machine is powered. With a business machine the language in use is loaded from a disk, either a floppy removable one, typically with a capacity of 360 kb or 1.2 Mb , or a fixed Winchester type with a storage capacity of $5-100 \mathrm{Mb}$. With this type of machine the operating system is also loaded when the machine is powered. It may carry out other tasks such as loading a program automatically. To do this the operating system looks for the presence of a particular file on the disk: the file will contain a list of commands which the operating system understands. With the DOS operating system this file is called AUTOEXEC.BAT; with CP/M the file is called PROFILE.SUB.

The final result of preparing a program which a computer can carry out is a list of numbers to be fed to the microprocessor. These numbers must be held in memory so that they are accessible while the program is being run. Clearly one would have to be something of a masochist to laboriously design and write a program in this form, as a list of binary numbers. The task would be prone to error, tedious and impossible for anyone, including the author, to understand at a later date. Even so, in the pioneering days of computers this was the way in which programs were written!

The purpose of a computer language is to convert these numbers, i.e. instruction bytes, into a more readable form. The basic instructions in binary number form are known as machine code. They can be considered as the "lowest" level of computer language - the level that a microprocessor understands directly. The next step up is to use a program called an assembler.

## Assembler

An assembler expects the user to write his program using meaningful mnemonics for the instructions. These are written in the form of a text file, using a word processor or editor - the latter is a form of word processor without some of the more advanced features found with word processors. The assembler is then run: it reads the file and converts the mnemonics into instruction bytes. Comments can be included in the text file - known as source code - to enable others to understand how the program works. The important point is that every mnemonic represents one instruction.

The source code shown in Table 1 is for a program to change the border colour on the screen of a machine that
uses an 8086/8088 microprocessor - an IBM or similar machine. There are several points of interest. It will be seen that the most-used mnemonic is MOV. This indicates that data is to be moved between registers in the microprocessor chip and/or memory addresses. The things that follow MOV are the destination and source of the data. Items in the right-hand column are comments that don't appear in the assembled code. The more observant will notice that it takes five steps to multiply a number by ten! The mnemonics used are decided upon by the microprocessor manufacturer: most producers of assembler software adhere to the standards. A note for frightened would-be programmers: this file has been shown purely as a matter of interest - it's programming at the deep end.

## High-level Languages

What we need is a language more like English, one that has single commands to carry out often used instructions such as the multiplication by ten just mentioned. The more akin it is to English the easier the program will be to understand - there's no point in making things unnecessarily difficult.

The most common high-level general-purpose language is BASIC. As mentioned in Part 2 BASIC is itself a program which may be stored on disk or in ROM. If on disk it must be loaded into RAM in the computer for a BASIC program to be written or run.

To write a program we type in lines of text, using words known to BASIC. The interpreter in the computer stores these words in shortened form by assigning a number to each word. This process is known as tokenising. We can then "save" (store) the program on disk or run it there and then. If neither is done the program will be lost at switch off, since it's stored only in RAM. Tokenising is not seen by the user, to whom the program consists of lines of text.

With BASIC each line can contain several instructions this is not the case with some languages. When the program is run the interpreter converts each instruction into the relevant machine code for the microprocessor used in the computer. One BASIC instruction can produce hundreds of bytes of machine code, so the use of a high-level language speeds up programming no end. The interpretation process slows down a program's running speed however. Two other factors that reduce the running speed are the syntax and error checking that the interpreter performs.

## Compiler

One solution to this speed limitation is the use of a compiler. The program is first written and tested in the normal way. It's then fed to the compiler which does all the syntax checking then turns the program into machine code once only instead of each time the program is run. In this way the file can be run without BASIC being present. Another advantage is that if the program is being sold commercially the source code is not released - it would therefore be very difficult to alter or "borrow" any of the program.

Many high-level languages are available only in compiler
form. Programs can thus be tested only by compiling then running them. Some errors will be thrown out by the compiler, but you can find you've written a valid program that doesn't do what you want.

Use of an interpretive language means that small sections of program can be tested quickly. In addition, most interpreters have a command mode in which instructions can be executed directly from the keyboard without being stored as a program. BASIC and dBase II/III are both of this type.

## Control Structures

Most languages have things called control structures. To explain this, a program is basically a series of instructions which are executed in order. Quite often however we want the program flow to change, depending on things like a keyboard input from the user. For example, suppose we have a program that prints a one-hundred page list of customers on request. We wouldn't want all this if we only wished to find out whether Mrs. Bloggs was in arrears with her rental. Clearly in these two cases the program has to be put into effect in different ways: a control structure enables this to be done, by changing the sequence in which program lines are called up.

Most if not all high-level languages use "condition" and "iteration".

## Condition

The condition control structure uses the words "if then else". If the result of the expression following the if is true, any commands on the same line, up to else, are performed. If the result is false, commands following else on the same line are performed. The else is optional. For example, we might have the line IF $X=2$ THEN PRINT " $X$ is two" ELSE PRINT " X is not two".

Some dialects of BASIC use a better if construction that's shared by other languages, as follows:

IF (expression)
commands
commands
commands
ELSE
more commands
more commands
more commands
END IF.
This is much better, as it allows for more lines of commands than the simpler version.

## Iteration

There are several varieties of iteration, which is used if we want to perform a group of commands several times. This is also known as looping. Consider a program that prints a message on the screen ten times (whatever use this might be!). It would be cumbersome to program the line PRINT "This is the message" ten times. Instead, we enter:

FOR J = 1 to 10
PRINT "This is the message"
NEXT J
which is known as a for/next loop. The action is that we use a variable, $J$ in this instance, though we could have called it $\mathrm{X}, \mathrm{Y}, \mathrm{Z}$ or FRED. The value of J starts as one, then the

Table 1: Source code example.

| Code (hex) | Mnemanics |  | Comments |
| :---: | :---: | :---: | :---: |
| 8A0E8000 | MOV | CL,(0080) | ;get tail length |
| 80 F902 | CMP | CL,02 | ;has tail at least 1 digit |
| 7230 | JB | 0139 | ;no tail |
| 80F903 | CMP | CL,03 | ;has tail 2 digits? |
| 772B | JA | 0139 | ;tail 3 or more digits |
| 740D | JZ | 011D | ;tail 2 digits - value |
| 8A168200 | MOV | DL,(0082) | ;move single digit up |
| 88168300 | MOV | (0083), DL | ;to same address as >9 |
| 33 CO | XOR | AX,AX | ;clear accumulator |
| E90F00 | JMP | 012C | ;process 1 digit |
| A08200 | MOV | AL, (0082) | ;get tens digit |
| 2 C 30 | SUB | AL,30 | ;ascii to int |
| D0E0 | SHL | AL, 1 | ;2n |
| 8AD0 | MOV | DL,AL | ;park it |
| D0E0 | SHL | AL, 1 | ;4n |
| DOEO | SHL | AL, 1 | ;8n |
| 00D0 | ADD | AL,DL | ;10n in al |
| 8A168300 | MOV | DL,(0083) | ;get units digit |
| 80EA30 | SUB | DL,30 | ; a to i |
| 00D0 | ADD | AL,DL | ;tens + units |
| BAD903 | MOV | DX,03D9 | ;port number |
| EE | OUT | DX,AL | ;bang! |
| C3 | RET |  |  |

message is printed. When the interpreter reaches the NEXT $J$ line it increases the value of $J$ by one and checks whether it has yet reached ten. If not the program reverts to the line following the FOR. Otherwise it continues with the line following NEXT. Thus all the lines between FOR and NEXT are performed a set number of times. Some languages don't have this feature.

Two other forms of iteration are found in many languages and in some versions of BASIC. They both rely on repeating a process until an event occurs instead of carrying out repetition a fixed number of times. There's a slight difference between the two structures, though it's not a vital one.

The first is the REPEAT/UNTIL loop. Here's an example, in BASIC:

## REPEAT

INPUT "Please enter your name, 999 to end";a\$
PRINT a\$
UNTIL a\$ = "999"
This is a nonsensical example but serves to illustrate the idea. The program will repeatedly ask for and print the name until you enter 999. The UNTIL line carries out the test. If the expression returns a false result the program goes back to the REPEAT. Thus the instruction(s) within the loop are always carried out at least once even if the condition is true to start with.

The other type of iteration differs in that the loop is repeated while a condition is true. Here's an example in dBase III:

## DO WHILE .NOT. EOF()

? name, addr,telno
SKIP
ENDDO
The test .NOT. EOF() simply checks whether the end of a file has been reached. SKIP moves through the file a record at a time. Don't worry about this: the significant point is that the test is carried out at the beginning of the loop instead of the end.

## CASE Structure

These control structures are mandatory for any computer language. There's another very useful one that's missing in BASIC. This is the CASE structure. Consider a program with a menu of choices from which the user has to chose a number or letter. With BASIC we would probably carry out an IF test on all the possible choices, or adopt some equally complicated method. The CASE structure eliminates this problem. Here's an example, in language $C$ :
switch (choice)
\{ case 1: command; break; case 2: another command; break case 3: yet another; break
\}

Much neater, isn't it? This particular section of code carries out different commands on the value of a variable
"choice". The word "break" is a part of language C to prevent execution of more than one command at this point.

F

## Threaded Interpretive Languages

Before closing this time we must mention another class of languages altogether. BASIC, C etc. are all procedural languages, i.e. the interpreter or compiler reads a list of instructions sequentially or in a sequence determined by a control structure. This other class of languages is called threaded interpretive - the best known example is FORTH. With these the language consists of a number of named routines known as words. You don't really write a program, but instead define new words in terms of existing ones, ending with a single word that executes the program. An application written in FORTH is really an extension of FORTH rather than something separate. We'll have more to say about FORTH next month, when we come to consider the suitability of these various languages for different applications.

## More Troubles

## Les Lawry-Johns

Well here we are again, tapping all the wrong keys and making a mess of everything. How the editor puts up with it I just don't know. Poor old Stan from SEME is also on the rocks. He can't do much driving, so we have to phone our orders in and make sure he gets the credit. One way or another we all seem to be up against it. Perhaps we're being tested. Like I was when this chap brought in a fairly new 14in. Fidelity portable, a CTV140 I think.

## The Fidelity Portable

It didn't want to work at all, and I didn't suspect the line output transformer as I would have done with the earlier ZX2000 chassis. When I had switched it off however I checked between the line output stage feed resistor and chassis. The reading was 20』. Probably the BY127 efficiency diode in parallel with the line output transistor (BU508A). I peered inside and failed to see it. Someone had taken it out and fitted it underneath, as I discovered when I withdrew the panel. On closer inspection I found that it was fitted the wrong way round. So I removed it and checked again. The low reading was still present. I was about to bawl at the line output transformer when I thought I'd better check the transistor first. It was the BU508A that was causing the trouble, so I apologised to the transformer and fitted a nice new transistor and put the diode in the right way round.

When I switched the set on again I was rewarded with a nice, clear picture. On fitting the rear cover I saw a label attached. Rapid Repairs. Oh well, that explained it all. These Rapid Repairs people have been going around lately causing havoc. Not Rapid Repairs, actually, but you know who I mean - don't you?

## Before I Forget

Time to thank those of you who've written in to wish me a rapid recovery from the brain shut-down that's been troubling me of late. I'd like to thank in particular Ken Muir of Maidstone. He suggested that a book called "Service with a Smile", illustrated by Giles and containing
some of my articles, ought to be published. Articles other than the Red Baron one. What was wrong with the Red Baron? Thanks to E.V. Hurran for the tip about vitamin E. Must try this. In reply to David Botto of Bournemouth, thanks, I've stopped taking the tablets - they seemed to make my head spin round instead of being hazy. Also John Wakely of SW19 - sorry I took so long to acknowledge your letter.

## Mr. Cole's ITT

Mr. Cole came in moaning his head off about his old ITT CVC5 I'd repaired before Christmas.
"It's gorn again. Now don't get me wrong, I'm not moaning, but it shouldn't have gone again so quickly, should it?"
"It depends on what's wrong with it."
"There's no sound. Here's the bill you gave me."
I looked at the bill. It said "replace the boost capacitor, $0.47 \mu \mathrm{~F} 1 \mathrm{kV}$, and test".
"That's got nothing to do with the sound" I said.
"Course it has. You did the set, didn't you? And it shouldn't have gone again so quickly."

So I told him to leave it with me to check over. I suspected the PCL86 audio valve but it turned out to be the loudspeaker. A new one put everything right and the sound was crisp and clear. I wrote on the bottom of the previous bill "fit new loudspeaker, previous one has given 15 years' service, $£ 5$ ".

When he came back he had a big smile on his face. I showed him his speaker and the bill and his smile faded.
"I'm not paying you any more money and that's that."
"O.k. Leave the set here and I'll sell it to get my money back."
"Not likely" he said as he tried to lift the set up. He couldn't, since I'd brought it in. "Help me get it to the car" he panted.
"Not likely" I said. "Pay your fiver or clear off."
So he paid his fiver and I picked up the set and put it in the car. If I'd known I'd have made it a tenner.

## Boozy Tessa

Tessa now has three saucers of sherry a night. Zeb won't drink but there's no doubt that Tessa's a drunkard. H.B. is on the wagon and says Tessa takes after her dad (you know who). All I have is a few scotches, only a few


# Wide-range Capacitance Bridge 

David Botto

Servicing TV sets and VCRs presents plenty of problems for the service engineer, not the least of which is checking suspect capacitors, especially of the smaller values. You may stock a comprehensive range but there always seem to be calls for the odd values that are not to hand. It's also a good idea to check new capacitors before they are soldered into the circuit - they have been known to be faulty!

Despite this, in the writer's experience relatively few TV/VCR service departments seem to possess an instrument that will measure capacitance accurately. You'll find that the capacitance bridge described in this articie will be in constant use on the bench, saving you hours of time and a lot of tension and frustration.

## Features

The instrument has five ranges, covering from 5 pF to $2,000 \mu \mathrm{~F}$. There's also a sixth range which is useful for accurately matching in value two or more resistors, capacitors or other components. Resistance ranges can be included if required. In addition a handy choice of two squarewave signal outputs is available for checking the sound circuitry in TV sets, VCRs, etc.

Because bench space is always at a premium the instrument has been designed for compactness: it measures $7.5 \times 4.33 \times 2.22 \mathrm{in}$. $(19 \times 11 \times 5.6 \mathrm{~cm})$. Battery operation was chosen for three main reasons. First so that the instrument can be carried easily for field servicing, without the need to hunt around for a spare mains socket in the customer's home. Secondly because the tester is more convenient to handle without trailing leads, and can easily be moved to any part of the workshop. And finally because mains operation would increase the size and weight considerably. Since it's in operation only when a measurement is being made the batteries enjoy a long life.

## Principle of Operation

Understanding the principles of operation helps in obtaining the best results from any instrument. The capacitance tester design is based on the well-known principle of the Wheatstone bridge. Fig. 1 shows the basic circuit. Resistors R1 and R2 have the same value. R3 and R4 also have the same value. With a d.c. voltage applied across points W and X , current will flow via the resistive potential divider chains R1/2 and R3/4. Since the voltages at Y and Z will be the same, no current will flow through the meter. If the ohmic value of R3 or R4 (or alternatively of R1 or R2) is altered the bridge will no longer be balanced and the meter's reading will deflect from centre zero to give either a positive or a negative reading. For example, if R 3 is reduced in value voltage V 3 will decrease and voltage V4 will increase, with voltages V1 and V2 remaining the same. The d.c. voltage at Z is now higher than that at Y. The bridge is unbalanced and the meter gives a positive reading - see Fig. 2. For the bridge to be balanced again the value of R1 must be decreased so that voltages V1 and V3 are exactly equal.

In Fig. 3 resistors R1 and R2 have been replaced with a linear potentiometer. Ratio 1 corresponds with R1 in Figs. 1 and 2 while Ratio 2 corresponds with R2. The unknown

R corresponds with R3 and the standard R with R 4 . The bridge circuit can now be used to measure resistance. An accurate resistor of known value is connected across terminals A and B . The resistor whose value is to be measured is connected across terminals C and D . If the two resistors are of equal value and the slider of the potentiometer is at track centre the meter will indicate zero voltage. If however the value of the unknown resistor differs from that of the known, fixed value resistor the potentiometer's slider will have to be moved up or down for the bridge to be balanced and give a zero reading on the meter. The value of the unknown resistor can now be obtained from the formula: unknown $\mathrm{R}=$ ratio $1 \times$ (standard $\mathrm{R} /$ ratio 2 ).
For example, suppose the value of the standard resistor is $10 \Omega$, ratio 1 is $80 \Omega$ and ratio $220 \Omega$. This gives us $80 \times$


Fig. 1 (left): Wheatstone bridge in the balanced condition.
Fig. 2 (right): Unbalanced Wheatstone bridge.


Fig. 3 (left): Wheatstone bridge with balance potentiometer.
Fig. 4 (right): Bridge for measuring capacitance.


Fig. 5: Basic circuit arrangement used in the Hunt's Capacitance Analyser.


Fig. 6: Circuit of the wide-range capacitance bridge.
$(10 / 20)=40$, i.e. the value of the unknown resistor is $40 \Omega$. With a set of standard resistors, resistance values can be measured precisely over a number of ranges. In practice the potentiometer is fitted with a scale that's calibrated in resistance values, doing away with the need for any calculations.

A capacitor has an ohmic reactance value Xc that's given by the equation $\mathrm{Xc}=1 /(2 \times 3 \cdot 14159 \times \mathrm{f} \times \mathrm{C})$, where f is the frequency and $C$ the capacitance value. By using an a.c. source voltage instead of a d.c. one the ohmic reactance of a capacitor of unknown value can be balanced against that of a known value capacitor in a bridge circuit - see Fig. 4. In this way the Wheatstone bridge can be adapted to measure capacitance.

## Hunt's Capacitor Analyser

In the 1940s and 50 s capacitance bridges such as the Hunt's Capacitor Analyser were found in most radio service departments. Earlier versions were built into a stout oak case with a removable lid: later versions had metal cases. A magic-eye tuning indicator was used to show bridge balance and there was a scale marked with resistance and capacitance values so that measurements could be read off directly. The basic circuit is shown in Fig. 5. Because the mains-derived a.c. source voltage is a very stable 50 Hz , resistors were often used as the capacitor standards. For example, at 50 Hz the reactance of a $1 \mu \mathrm{~F}$ capacitor is $3,180 \Omega$ and that of an $0 \cdot 1 \mu \mathrm{~F}$ capacitor $31,300 \Omega$. These instruments usually had a range of $0.0001-100 \mu \mathrm{~F}$ and about $5 \Omega-10 \mathrm{M} \Omega$, which was quite adequate for servicing the valve radios of the time. These old bridges still give good service in a few workshops.

## Circuit Description

The capacitance bridge that forms the subject of this article uses two inexpensive i.c.s. Fig. 6 shows the complete circuit. A sinewave voltage is generally used to power a test bridge but experiments have shown that almost any type of a.c. waveform can be used. This circuit employs a 555 timer i.c. to provide a squarewave output. The arrangement has
the advantage of being simple, few components being required. The 555 chip's frequency of oscillation is determined by the value of $\mathrm{C} 2(0 \cdot 1 \mu \mathrm{~F}), \mathrm{R} 2(1 \mathrm{k} \Omega)$ and the setting of VR2. With SW2 in the open position the frequency is 146 Hz . When SW2 is closed R3 is connected in parallel with VR2 and the frequency is increased to approximately 456 Hz .

The squarewave oscillator's output is applied to the primary winding of transformer Tl , whose secondary feeds the bridge. VR1 is the calibrated balance potentiometer while capacitors C6-C10 are the standards against which the capacitor under test is balanced. Since resistance values can be accurately measured with a digital multimeter, resistance ranges were not included in the prototype. It's simple to add resistance ranges to the bridge if required. Use close-tolerance 0.5 W resistors of $100 \mathrm{k} \Omega, 10 \mathrm{k} \Omega$, $1,000 \Omega, 100 \Omega$, and $10 \Omega$ as the standard balancing resistors. Fig. 7 shows the additional circuitry. The range covered is from $1 \mathrm{M} \Omega$ down to the resistance of five inches or less of 22 s.w.g. single-strand connecting wire! This last range is useful for checking the windings of low-resistance coils for shorted turns.

VR1 is a linear, wirewound $10 \mathrm{k} \Omega$ potentiometer. Its balance point is detected by the LM386 audio amplifier which drives a 2 in ., $8 \Omega$ loudspeaker. The balance signal from the bridge is coupled by C 5 and the preset gain potentiometer VR3 to pin 3 of the LM386 chip. When the correct balance point has been found there will be no output to the LM386 and thus no sound from the loudspeaker. The capacitance of the component being tested is then read directly from the calibrated scale.

The small control VR4 is the power factor control. It's used on the two higher capacitance ranges to obtain a sharper null balance and to indicate the power factor of the capacitor being tested.

Two separate battery supplies are provided, one for the 555 oscillator and the other for the LM386 audio amplifier. This prevents unwanted coupling upsetting the accuracy of the bridge balance. Another advantage is that the batteries enjoy a longer life. It's important that the earth sides of the two supplies are not connected together.

Jack socket J1 is the output for the squarewave test
signal: in this application VR1 acts as a gain control. J2 enables a scope to be connected to serve as a null point indicator, with the sound cut out.

## Construction

Construction of the instrument is straightforward and the parts required are all readily available. The accompanying photographs show the finished appearance and internal layout of the tester. A plastic case is used - don't use a metal case because this could cause problems as a result of internal capacitances. A Tandy de luxe project case was used for the prototype, catalogue no. 270-224. Any similar plastic case is suitable.

Fig. 8 shows the recommended layout for the controls and the terminals - the positions were chosen to avoid unwanted circuit coupling. The i.c.s are mounted on two separate experimenters' "perfboards", Tandy catalogue no. 276-150. Each perfboard is held in place securely with double-sided sticky pads (these are available from all good hardware shops). Use of these perfboards makes it unnecessary to etch your own PCBs, thus saving a great deal of time and effort. Make sure that the two battery holders are fitted to the bottom of the case securely - use a little Blue-Tack.

The accuracy of the bridge depends on the tolerance of capacitors C6-C10. C6 is a 100 pF silver mica type accurate to $\pm 1$ per cent. C 7 and C 8 are good quality capacitors that were found to be quite close to their stated values. C9 and C10 are small, 25 V working electrolytics. Don't solder C9 and C 10 into circuit until you've set up the other ranges.

T1 is an RS Components type T/T3 which happened to be in our stock (it's no longer listed). Any small audio driver transformer is suitable - don't use an output transformer. Twist together the two insulated wires from the transformer's secondary winding to the ends of the bridge arm.

The miniature on/off switch SW1 is a double-pole, double-throw type with three positions (on/off/on). Mark the off position clearly so that the instrument will not be accidentally left switched on while not in use.

Fig. 9 shows a full-sized calibrated scale which you can trace or copy, saving yourself a lot of work. The best material to use for this is white Bristol board, which should be available from your local art shop. In order to use the


Internal layout of the prototype. For external view see photograph on front cover.
scale as shown it's essential to fit the specified RS Components $10 \mathrm{k} \Omega$ linear control (type 173-237). After final testing, cover the scale with a piece of stiff, clear plastic approximately $1 / 16 \mathrm{in}$. or 1.5 mm thick to keep it clean and free from accidental damage.

It's also an idea to fit four small self-sticking cushion feet to the bottom of the case. This will stop the instrument sliding about on the bench.

## Setting up and Calibration

There's nothing difficult about setting up the capacitance bridge. Before you do so ensure that all the wiring and connections are in order. Then set the two preset controls VR2 and VR3 to their mid-positions and the balance control VR1 to mid-track. Turn the range switch SW3 to the balance position and SW2 to its 146 Hz setting.

Connect the batteries and, if you've a frequency counter, plug it into jack socket JI via a 10:0 isolating probe. Turn the power switch SW1 to the output position and if all is well the counter should give a reading in Hz . Adjust VR2 for an output at 146 Hz . Close SW2 and the counter should give a reading in the region of 456 Hz - this is not critical. Disconnect the counter. If you don't have one, set SW3 to position C, the power switch to on and adjust VR2 for a pleasant low-pitched buzz from the loudspeaker. Closing SW2 should produce a higher tone.

Note that a single scale is used for all five capacitance ranges. To calibrate the bridge, connect an accurate $0 \cdot 1 \mu \mathrm{~F}$, 250 V working capacitor across the test capacitor terminals and set SW3 to position C. Rotate VR1 until you find the position that gives the minimum output from the loudspeaker (use the 456 Hz setting). VR3 should be adjusted for sufficient, not excessive, sound from the loudspeaker. When the null/minimum sound position has been found, set VR1's knob exactly to the centre position (10) on the scale. This should be at the control's mid-track position. If you now measure other capacitance values you should find that the accuracy of the scale is already quite good.

For correct calibration however you'll need a range of accurate capacitors with values between $3 \cdot 3 \mu \mathrm{~F}$ and $0.002 \mu \mathrm{~F}$ - see Table 1. Notice that with the exception of the centre 10 position the calibration lines shown in Fig. 9 don't quite connect to the centre of the scale. Using range C , connect each test capacitor in turn to the test capacitor terminals and adjust VR1 for minimum sound. After each check link the calibration line to the edge of the scale - see Fig. 10. For this range the scale numbers have to be divided by one hundred: the centre 10 represents $0 \cdot 1 \mu \mathrm{~F}, 47$ stands for $0.47 \mu \mathrm{~F}$, etc.

It's not necessary to calibrate all the other ranges once the C range has been calibrated correctly. It's best however to check the 5 pF and 10 pF balance points on the A range, using 2 per cent tolerance silver mica capacitors for the purpose.

The balance scale should be read as follows.
Range A: Centre scale 100 pF . Scale numbers times ten (read in picofarads).
Range B: Centre scale $0.01 \mu \mathrm{~F}$. Scale numbers in $\mu \mathrm{F}$ divided by 1,000 .
Range C: Centre scale $0 \cdot 1 \mu \mathrm{~F}$. Scale numbers in $\mu \mathrm{F}$ divided by 100 .
Range D: Centre scale $10 \mu \mathrm{~F}$. Scale reads in $\mu \mathrm{F}$ directly. Range E: Centre scale $100 \mu \mathrm{~F}$. Scale numbers in $\mu \mathrm{F}$ multiplied by 10 .

Two small 25 V electrolytics are used as the standard capacitors in the $D$ and $E$ ranges. You can use the


Fig. 7: Extra components required to add resistance ranges. SW3/SW3a comprise a two-pole, six-way switch


Fig. 8: Suggested drilling details for the case top. Hole sizes depend on the components used.

Table 1: Components for precise calibration.
Capacitors: $3.3 \mu \mathrm{~F}, 1 \mu \mathrm{~F}, 0.47 \mu \mathrm{~F}, 0.33 \mu \mathrm{~F}, 0.22 \mu \mathrm{~F}, 0.15 \mu \mathrm{~F}, 0.1 \mu \mathrm{~F}$, $0.068 \mu \mathrm{~F}, \quad 0.047 \mu \mathrm{~F}, \quad 0.033 \mu \mathrm{~F}, \quad 0.022 \mu \mathrm{~F}, ~ 0.015 \mu \mathrm{~F}, ~ 0.01 \mu \mathrm{~F}$, $0.005 \mu \mathrm{~F}, 0.002 \mu \mathrm{~F}, 10 \mathrm{pF}$ (silver mica $2 \%$ ), 5 pF (silver mica $2 \%)$.

Use range C. Divide scale numbers by 100.
Resistors: $3 \cdot 3 \mathrm{k} \Omega, 1 \mathrm{k} \Omega, 470 \Omega, 330 \Omega, 220 \Omega, 150 \Omega, 100 \Omega, 68 \Omega$, $47 \Omega, 33 \Omega, 22 \Omega, 15 \Omega, 10 \Omega, 5 \Omega, 2 \Omega$.

Use range B. All resistors should be close-tolerance types - gold 5\% or better.

## Components List



R2
VR1 10k 1 W linear wirewound, RS173-237
VR2 470k $\Omega$ miniature horizontal preset
VR3 10k $\Omega$
VR4 $1 \mathrm{k} \Omega$ miniature horizontal preset miniature linear panel-mounting control
C1 $470 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ tubular electrolytic
C2 $\quad 0.1 \mu \mathrm{~F} \quad 150 \mathrm{~V}$ tubular
C3 $10 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ tubular electrolytic
C4 $\quad 0.1 \mu \mathrm{~F} \quad 150 \mathrm{~V}$ tubular
C5 $\quad 0.05 \mu \mathrm{~F} \quad 150 \mathrm{~V}$ tubular
C6 $100 \mathrm{pF} \quad 1 \%$ silver mica, RS 124-780
C7 $0.01 \mu \mathrm{~F}$ polystyrene, RS 113-409
C8 $\quad 0.1 \mu \mathrm{~F} \quad 100 \mathrm{~V}$ epoxy cased ceramic plate, RS 125-733
C9 $10 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ electrolytic
C10 $100 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ electrolytic
C11 $\quad 0 \cdot 1 \mu \mathrm{~F} \quad 150 \mathrm{~V}$ tubular
C12 $\quad 0.1 \mu \mathrm{~F} \quad$ 150V tubular
C13 $10 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ tubular electrolytic
C14 $1 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ tubular electrolytic
C15 $100 \mu \mathrm{~F} \quad 25 \mathrm{~V}$ tubular electrolytic
IC1 555 timer
IC2 LM386 audio amplifier
T1 RS T/T3 or similar audio driver transformer
SW1 Miniature DPDT on-off-on toggle switch. Tandy 275-620 or similar
SW2 SPST microminiature toggle switch. Tandy 275624 or similar
SW3 Six-way two-pole rotary switch. RS 327-658 or similar
Two 3.5 mm jack sockets
Miniature $8 \Omega$, 2in. loudspeaker
Two penlight battery holders each to hold four batteries, plus snap-on connectors. Tandy 270-383 or similar
Two perfboards. Tandy 276-150 or similar
Plastic case, e.g. Tandy 270-224
Self-sticking cushion feet. Tandy 64-2346 or similar
Additional components for resistance ranges:
R4 $10 \Omega$ RS $149-616$, R5 $100 \Omega$ RS $149-644$, R6 $1 \mathrm{k} \Omega$ RS
149-694, R7 10k RS149-818, R8 100k $\Omega$ RS 149-925
SW4 SPDT miniature switch
accurately calibrated C range of the bridge to measure a number of $10 \mu \mathrm{~F}$ and $100 \mu \mathrm{~F}$ electrolytics, selecting two that have exactly the required values.

## Resistance Ranges

If resistance ranges are to be added to the bridge you'll find that the additional figures required on the balance scale will be the mirror image of those used for the capacitance ranges. For proper calibration a range of close-tolerance resistors is needed - see Table 1. Check their accuracy first with a digitial multimeter. The resistor to be checked is connected across the test capacitor sockets. Draw the ohms scale as shown in Fig. 9, with the same numerical values but as a mirror image. Mark it out on the outer edge of the capacitance scale, leaving room for calibration link lines. It's easily calibrated by selecting range B. Connect each of the resistors listed in Table 1 to the test sockets in turn. Adjust VR1 for minimum sound.


Fig. 9 (left): Full size calibrated scale for the bridge.
Fig. 10 (right): The 'link" method of calibration, which allows for component variations.

As you do so, mark in their numerical values - divided by ten - on the resistance scale.

If fitted, the resistance ranges read as follows.
Range A: Centre scale $10 \Omega$. Scale reads directly in ohms. Range B: Centre scale $100 \Omega$. Scale numbers multiplied by ten.
Range $C$ : Centre scale $1 \mathrm{k} \Omega$. Scale numbers multiplied by one hundred.
Range D: Centre scale $10 \mathrm{k} \Omega$. Scale numbers multiplied by one thousand.
Range E: Centre scale $100 \mathrm{k} \Omega$. Scale numbers multiplied by ten thousand.

## Circuit Varnish

After calibration it's a good idea to apply a thin coat of circuit varnish to all solder tags, joints, etc. Don't spray the varnish on - use a small brush. This will help to ensure reliability in the long term.

## Use

A capacitor to be tested can be connected to the test capacitor terminals directly or via test leads. In the latter case note that with small picofarad value capacitors the presence of the test leads might result in an incorrect reading.
Make sure that a capacitor is discharged before connecting it to the test terminals - especially in the case of large-value reservoir and smoothing capacitors.

Select the appropriate range with SW3, turn the scale pointer until the minimum volume note is heard from the loudspeaker, then read off the value from the scale. A balance point at the extreme left of the scale indicates a short-circuit while a balance point at the extreme right indicates an open-circuit.

A convenient feature of the bridge is that an oscilloscope can be connected to jack socket J2, enabling the balance
point to be observed visually with the sound cut out.
When checking electrolytic capacitors, note the normal position of the power factor control for the sharpest null balance. The term power factor relates to losses and leakage in a capacitor. If you find that the power factor control has to be turned well beyond the normal position the capacitor being tested is leaky.

You'll soon get used to reading the power factor and scale pointer figures. Measurements can be made using either oscillator frequency. The lower frequency is best for large-value capacitors, the higher frequency for picofarad values.

Some capacitors can be checked in circuit - disconnect the equipment from the mains supply first! In many cases however the measurement will be affected by other components in the circuit. The easiest way to isolate an in-circuit capacitor is to unsolder one of its leads and keep this clear of the printed circuit. Check the capacitor via your test leads, then resolder the lead if the capacitor tests good. This saves time and the frustration of struggling to replace capacitors in positions that are hard to get at.

The balance range is very helpful when you need matched components and don't have them to hand. Two components such as capacitors, coils, resistors, etc. can be matched exactly. One component is connected to the test capacitor terminals, the other to the balance terminals. When two components are matched correctly the bridge will balance with VR1's pointer at the half-way scale mark.

The two squarewave signals available are extremely useful for checking audio circuitry. Turn SW1 to the output position and plug a screened lead into jack socket J1. The lead should be at least a metre long so that the capacitance bridge's unscreened case is well away from the equipment under test. VR1 now acts as a volume control. If you connect a scope to J1 you'll see the squarewave output increase and decrease as the control is turned.

As you become accustomed to your capacitance bridge you'll find new uses for it.


# Letters 

## SERVICING INDUSTRY CHANGES

From the letters in your January issue it appears that many readers are aware of the changes taking place in the TV and video servicing industry. While it will take another couple of years for the new order of servicing to be fully established, anyone in the industry who doesn't recognise the changes will be left out in the cold.

Martin Blake's letter is significant in drawing attention to the new attitude being taken by many manufacturers. Rather than adopt the negative approach of rejecting Grundig products, he ought to take the initiative and contact Grundig and its dealers in his area. They may be all too willing to off-load some of their servicing. Reputable manufactureres are willing to support servicing organisations that are, in turn, prepared to invest in equipment and training. Grundig and JVC in particular do so. But why this new attitude and what does it involve?

Much of the TV and video equipment produced by the major manufacturers today is of very advanced design, incorporating complex microcomputer routines that make it possible to have access to internal functions via the microcomputer. Also we are seeing the first examples of equipment with field stores to give picture-in-picture and other features. There are Grundig VCRs with the ability to override emergency fault routines, JVC camcorders with the same facilities, Finlux TV sets with internal adjustments via the remote control handset, and both Sanyo and Toshiba VCRs with field stores. Without backing from the manufacturers, servicing this type of equipment will be beyond the service engineer. He won't stand a chance of repairing it and will end up returning the equipment to the customer - with a high probability of additional faults.

You may say it's all right for Steve because he has Grundig Pete to talk to. But there have been times when we've both been confused over a simple VS180 and have not known, after someone has had a go at it, in which of two panels or three switches the fault has lain.

Say that you are faced with a new Panasonic NVM5 and the contrast of the picture is slowly changing up and down again. Nowhere does it mention that this is normal in the high-speed shutter mode, or that in a certain new VCR the slow-motion tracking control is now the same knob as the normal one. It gets to the point where the operating procedures can be as confusing as the internal circuitry! When faced with a new, faulty product for the first time you may not even be able to operate it, let alone repair it.

There are other points in the same issue. The man with the battered Marina van who finds it less casy/viable to carry out field repairs, so that more have to be done on the bench. "Ex-valve set dabbler" complains that the magazine has changed from DIY towards the trade. There's no such thing as DIY vidco repairs. You can't swap syscon or LSI servo chips like you could valves. For one thing they don't plug in, and for another without very good quality soldering/desoldering equipment the print will be damaged and that will write off the PCB and the VCR or TV set for good.

There will be plenty of opportunities for enterprising engineers who are prepared to spend money on equipment and time on training to set up regional service
centres for their local dealers, because as time progresses and domestic electronic equipment becomes more complex most sales-orientated dealers will not be able to cope with their own servicing. Think about it . . .
Steve Beeching,
Barnby, Newark.

## GRUNDIG'S NEW POLICY BACKED

I was amazed to see criticism of Grundig's new service policy when to me, a Grundig dealer of long standing, the company is now getting it right. It would take a full article to go into this in the depth it deserves, but I would sum it up as a policy of looking after those who look after you. If we look after the people who buy from us, and Grundig looks after us, that's as much as anyone is entitled to expect - but consider yourself fortunate if that's what you get.

In the past we had difficulty in getting through to Technical Information and Spares, mainly due to nondealers taking up telephone time. Grundig has made every effort to give its dealers information through service courses, technical bulletins, etc., but there were still occasions when we had a query that couldn't be dealt with because Technical's phones were engaged while they answered simple queries from non-dealers because they hadn't been on courses etc.

Your correspondent refers to people who move from one district to another being unable to obtain service. There are bound to be some drawbacks with any system, but as more dealers become aware of the Grundig dealer support I would expect it to be only a matter of time before there's a Grundig dealer in every town.

What I am amazed about is that no one appears to be complaining about Philips. Its back-up is so poor as to be virtually non-existent. The company closed down all its service departments and appointed "specialist dealers". We approached one regarding an under guarantee compact disc player. They told us they knew no more than we did about these machines, so they might have to spend some time on the fault. Philips allow only $£ 10$ so we’d have to pay any additional costs ourselves. What sort of back-up is that? The profit margins rarely allow a 25 per cent mark up, so in the event of a problem arising the chance of a final profit is remote. We changed from dealing direct to dealing through a wholesaler, who at least looks after any under guarantee items.

When Philips sold its last V2000) series VCRs the company didn't run service courses but trained four/five engineers who operated from Manchester. Any faulty machines were sent there via Securicor. This was a shortsighted arrangement, since at the end of the day the dealer knows nothing of these machines - and now's the time one can expect them to start coming in for service. Perhaps Philips would care to tell us whether the bods in Manchester are still slaving away, or provide some helpful advice so that dealers can cope with any future problems. R. K. Caley,
R.K. Electrics of Ilfracombe.

## OBTAINING A VARIABLE AC SUPPLY

I read with great interest Albert Hitching's article in the November issue, describing a versatile bench transformer. During the course of my work as an electrical breakdown technician I visited a firm of hose equipment specialists who were electroplating small parts to fit on the end of high-pressure hoses. They were using a small battery
charger with lamps connected in series to set the current as required. They wanted a better method of doing this, but as always didn't want to spend much money. I fitted an MK dimmer unit in series with the primary winding of the transformer and chucked out all the lamps. MK states that some of its dimmers can be used to control transformer loads in this way provided a $470 \mathrm{k} \Omega$ resistor is connected across the transformer's primary winding.

I've since used this idea to control a bench isolating transformer. This gives me any voltage I require and also a nice slow start to anything I'm working on. With fuses fitted where needed this seems to be a safe arrangement, though maybe some of your contributors who write with all those letters after their names will pull the idea to bits. But that's life - it works!
S. J. Searle,

Colchester, Essex.

## A NEW TYPE OF COWBOY

The cowboys have struck in Huddersfield town centre. Not the normal rusty van/run-down shop type of cowboys but a new breed - fast sales talk and expensive shop type cowboys. They are causing considerable distress to all respectable dealers. Their prices seem to be reasonable to Joe Public, but what they are selling should have been scrapped a decade ago. The main point that's causing concern is that they are advertising these second-hand goods as bankrupt stock, with no guarantee.

I've met countless victims of these con-merchants, people who have bought a TV set or VCR that broke down several days later and were told by these "professionals" that they would not carry out the repairs required. When outraged customers take a stronger line and talk about consumer protection they are told where they can go.

I've tried to repair several of these bankrupt stock items and have been amazed that they worked in the first place. People like this not only upset their customers but tend to give the impression that all second-hand TV firms are the same. Something must be done to stop these cowboys. They might be in your town next.
J. P. Roebuck, Britannia Electronic Industries, Huddersfield.

## REPAIRS TO VIDEO CASSETTES

Harold Peters provided some useful tips in his article on servicing VHS cassettes in the January issue. Here are some more, based on my experience.
(1) Spools used by TDK and Maxell/Hitachi are incompatible with JVC shells, though JVC's spools are compatible with TDK and Maxell/Hitachi shells. I've found that TDK and Maxell are in all respects compatible, assuming the format to be the same of course! Note that JVC makes tapes for Ferguson, Baird, Thorn, Akai, ITT and Kodak.
(2) The best method to adopt when removing the leader retaining wedge from the spool is as follows. Turn the spool over (white side up), insert a long ball-point pen tip into the small hole and push the wedge out. Under no circumstances prise the wedge out with a screwdriver blade - the brittle plastic is likely to fracture or break, causing further heartache.
(3) Assuming that the procedures described in the article and above are carried out no problems should arise. It would be far better to discard faulty cassettes, but what about irreplaceable recordings? In serious cases one
would do better to have the recording(s) copied on to a fresh tape before discarding the original - current HQ VCRs are capable of producing excellent copies.

Finally, never use unbranded or unheard-of brands.
In passing, I'd like to sympathise with the man in the battered Marina van. At the ripe old age of twenty two I often wish I'd been this age in the days of faulty valves and open-circuit mains dropper sections. As a toddler in the sixties I recall frequent visits from the NE Co-op engineer who used to attend the family set (a Defiant 9 961U I seem to recall) which frequently gave trouble. Far from screaming, I found the set and its interior fascinating, though the engineer preferred me to be out of the way. But that's another story!
Brian Renforth,
Newcastle-upon-Tyne.

## BETTER CABINETS WANTED

The things that make servicing more complex are switchmode power supplies, diode-split line output transformers and cmos circuitry. Are we paying too high a price for the increased efficiency and power saving that these techiques offer? Personally I wish that manufacturers would revert to more robust cabinets instead of improving the works inside. It's paradoxical that whilst the technology inside TV sets has progressed by leaps and bounds the cabinets, now made of flimsy plastic, have never been worse.
K. J. Freeby,

Plymouth, Devon.

## STILL PRACTICAL!

Your correspondent "ex-valve set dabbler" is right to point out that Television is read more by the trade these days. But take a look at the publications they used to read. The technical ones require fluent hexadecimal and total silence to be understond, while with the trade papers it's difficult to tell where the copy ends and the ads begin.

As for not caring for the enthusiast, I personally in the last year or so have described a method of finding Eutelsat-1 using a milk straw, school protractor and string; a way of videotaping teletext using a single transistor; and how to extend tape life using a pencil, rubber and scissors! F.J. Camm will pause from repairing his celestial twovalver to look down on us benignly for that! Having contributed to the magazine under F. J. Camm, Ray Street, Norman Stevens etc. I personally prefer the current style - the editor uses his blue pencil to add topicality rather than to delete anything with the slightest twinkle of humour in it.

The Western Brothers used to say "there's only a few of us left" (remember?). Les is on his tablets, and so am I to a lesser extent - such is the cut and thrust of fixing today's unfixables. So how about it "ex-valve set dabbler" and others like him? You must have come across something that will be of interest to us all. Simply write it up so that the chap going home on the train after a hard day will enjoy it. I can promise you the sky won't fall in!
Harold Peters,
Lowestoft

## NOT DETERRED

I agree with "ex-valve set dabbler" - your magazine is not catering for the DIY man. I don't care what the trade people think, I for one am going to have a go - at anything. It's about time we had a magazine that did
something for us like Practical Television used to do, concentrating on a particular set with complete circuit diagram - over two or three issues if need be.
Geoff Hope, electrician,
Guisborough, Cleveland.

## DABBLERS CONDEMNED

It's a good thing "ex-valve set dabbler" didn't sign his name. If I got hold of him I'd be likely to punch him on the nose! How many times have I seen fuses covered with silver paper, soldering like arc welding - even VCR heads cleaned with Germolene. And you can bet that the people who bring these sets in will tell you "it was all right last night". If only my life was centred around an endless supply of replacement panels: maybe I could pack up at dinner time and go home.

After spending three years at technical college (and I'm still not finished) I found that letter an insult. A good technical background and years of experience are needed to do this job properly. That's why some people are dabblers and others experts.
From a very angry Peter Goodman,
ex-Kettle repairman,
Corby, Northants.

## ENTHUSIAST REPAIRS

I would like to echo the comments made by "ex-valve set dabbler". There has been much talk in your columns recently about service charges, trade-only suppliers and rip-off merchants. Although professionally involved in electronics, as a circuit designer in the computer industry, I repair TV sets for friends simply because I find it a fascinating field, not for any financial benefit. One advantage is that I can mend sets which those in the trade wouldn't consider to be worthwhile. I've spent many hours slaving over a G8 or an A823 that would otherwise end up on the scrap heap. I've even repaired stock faults on sets that respected dealers have said were beyond repair. I'm not kicking the trade, but there are still clearly a lot of hobbyists like myself who repair things purely out of interest and to help others.

A week or so ago I found myself reading some of your issues from the seventies. They were refreshingly interesting, full of practical tips, and there were many more component advertisers. The magazine seems to be trying to cover too wide a field today, the practical TV side being displaced by articles on VCRs (a field that's not really suited to the hobbyist), computers (widely covered elsewhere) and how to run a business. The advertisements also have a strong trade bias - I personally have no use for Glls by the bucketfull.

I suggest you cast your eyes back to what the magazine was saying ten years ago. I agree that times have changed, but would hazard a guess that the number of hobbyists has changed little since then. You should still cater for our needs.
D.W. Sergeant,

Bracknell, Berks.

## A CLUB FOR DABBLERS?

I'd like to meet "ex-valve set dabbler" so that we can get dabbling together. You see I'm an unemployed faultfinding electrician with experience of many types of equipment. My hobbies are amateur radio and electronics.

Last year I managed to get a job for three months, but when they found out I was diabetic I was given a week's notice. Since then I've been unable to get even an interview because of my health and have turned to doing whatever repairs I can for others in a similar impecunious position to my own. Trying to get spare panels etc. when you cannot afford them is a tricky business (has anyone a spare power supply for a G11, the same plus a decoder/ signals panel for a G8?).

What I want is to do TV servicing in a workshop. If anyone is interested I'll send my details, c.v. etc. It would be an idea if dabblers and those who are unemployed but have some technical knowledge could get together to form a club. Let's all dabble together!
Ian Ruddock, G8NCZ, 54 Woodcroft Avenue, Stanstead Abbots, Ware, Herts SGI2 8JQ.

## MODIFICATION WANTED

I wonder if any reader could suggest a modification to the Philips 2023 VCR to override the three-four minute automatic switch off (unless a deck function is in use) so that the machine can be used to remote control a nonremote control TV set without the annoyance of loss of program cvery few minutes?
R.W. Silver,

Glasgow.

## SPARE PARTS QUERY

Does anyone know of a spare parts supplier for Silver products? Tech-Semco used to be able to provide spares but is no longer in the spares business. Perhaps one of your readers might be able to help?
Simon Kelly, JKL Electrix,
Newcastle, Co. Down.

## MMDS AND IRISH CONDITIONS

The MMD system described in the November issue is ideal for short-range links between line-of-site reception points, e.g. for extending a cable system to an adjoining town without the need for costly underground cabling. Here in Eire there's a need to extend multichannel reception of UK signals to smaller urban areas by retransmission from local clevated sites. The problems relate to controlling the spread of reception, achieving adequate financial arrangements and dealing with copyright requirements.

In the case of small, isolated towns and rural areas however MMDS is likely to be a costly business and distribution at $2 \cdot 5 \mathrm{GHz}$ could present many difficulties.

I note with interest reference to Canadian firms in the MMDS article, and recall some previous Canadian efforts at signal distribution in Eire. A decade ago Waterford City Cabling had Canadian experts who chose the wrong reception site and the wrong UK u.h.f. source (from S.W. England instead of S.W. Wales), then spent a couple of years "experimenting" whilst the long-suffering viewers waited for satisfactory signals. To this day some nearby unauthorised rebroadcasters provide reception that's superior to the cable system. In the case of Cork City more Canadian experts "experimented" and chose the wrong mountain range for reception, despite the advice of many Irish experts. Subscribers are of course paying for these past mistakes. I hope that more experiments by Canadians, this time in the MMDS field, are not going to be foisted on Irish viewers.

Personally I'd prefer to see redistribution at u.h.f. with simple encoding. The use of well-tried u.h.f. technology and local expertise could provide signals at a very reasonable cost - you wouldn't need a multimillion pound organisation to operate such a system. MMDS can be received on a line-of-sight basis only, and linear output amplifiers with outputs in watts would be very expensive. The same power levels at u.h.f. would provide a much more effective TV coverage in Irish rural areas. I can't see the merits of using untried, expensive s.h.f. systems when practical u.h.f. systems would suffice. In the USA and

Canada MMDS operates from high hill sites overlooking wide open expanses of flat terrain - perhaps line-of-sight up to 100 miles. The geographical conditions in rural Ireland are quite different.

Any proposed developments should be given very careful research and experimentation before we see a rush of microwave dishes on our hillsides. Perhaps DBS TV will come to our aid before MMDS. Time will tell.
Des Walsh,
Carrigaline
Co. Cork

# A Low-cost TVRO Installation 

## Part 1

Roger Bunney

There is very little satellite TV reception at present amongst DX-TV enthusiasts. Several have made their own equipment however while others have invested sums of hard-earned cash in commercial receiving installations equipment from Connexions seems to be favoured. It's a hard economic fact that even a basic system to traverse the heavens, using an azimuth/elevation mount, is likely to cost you $£ 650$ plus. For a motorised system with computer memory and an up-market receiver the figure rises to
 cables. In addition you might need a line amplifier.

As regular readers will know, I've been involved in terrestrial DX-TV reception and experimentation for many years. Much of my equipment has been home built, and spending large sums of money on equipment goes against the grain. Though so few DXers are active with satellite TV reception I've for some time felt that this subject should be given greater attention. To encourage others I decided to see what could be done. The main aims have been to minimise the cost of the exercise, incorporating homemade innovations where possible, while obtaining results comparable to those provided by a more up-market system. Cost saving has been achieved by accepting a degree of operational inconvenience that would probably put the normal domestic viewer off. It was also felt important that any DIY aspects should be repeatable by others who have little or no knowledge of TVRO installations - though the enthusiasm characteristic of UK TV-DXers has been assumed!

## Selection of Equipment

When you look through the advertisements in Television and the various video magazines you'll see quite a wide range of TVRO equipment on offer, much of it expensive. What we are seeking is the cheap gear, at the lower end of the market, which means manual receivers, patio mounts and the domestic packages aimed at the DIY or "pub" market. I decided to opt for the cheapest - a 90 cm dish with a head unit having a noise figure of $1 \cdot 8-2 \mathrm{~dB}$ and a manual receiver. The head unit picks up the signal collected by the dish and converts it from s.h.f. to a lower frequency (the first i.f.) for feeding to the receiver unit itself. It's the convention today to refer to the electronic part of the head unit as an LNB (low-noise block), so we'll use this term from now on. In the patio mount field it's unlikely that you'll find dishes with a diameter of more than about a metre.

With the low-power satellites we're aiming to receive a

Y cm dish provides an LNB input that borders on the marginal, so a very low noise system is essential. At under $£ 500$ retail typical performance figures are gain of around 55 dB with a noise figure of 2 dB or lower. If you can go for a 1 m dish, so much the better. The latest LNBs use HEMFET technology, with noise figures of less than 1.5 dB - but you pay for this enhanced performance! The system I've put together uses a 90 cm dish and an Echosphere LNB feeding, via RF125 u.h.f. coaxial cable, an Echosphere SR1000e receiver. Having sounded out the market for possible sources of supply - not an easy task - I decided to purchase the equipment from North East Satellite Systems of Cropton, North Yorkshire. John Standen of North East Satellite Systems is noted for his expertise in the commercial satellite market and I feel that his company's track record gives assurance should any problems arise.

I decided that use of a polarotor for remote change between vertical and horizontal polarisation was unwise since it introduces a loss approaching 3 dB . This is unacceptable with a dish of less than 1 m (we're not talking about DBS reception!). It means that the LNB/feedhorn assembly will need to be physically rotated to suit the polarisation of the required transponder downlink. Inconvenient - but a financial saving! I bought an adjustable scalar ring assembly since this allows you to "tune" the head for optimum signal pickup from the dish. Doing this can provide an improvement of $0 \cdot 5-1 \mathrm{~dB}$. The LNB is fed with an input at $10 \cdot 9-11 \cdot 7 \mathrm{GHz}$ and provides a downconverted i.f. output at $950-1,750 \mathrm{MHz}$.

Having decided to buy the Echosphere units a cheque was sent off. Shortly afterwards two large cardboard boxes arrived . . .

## The Patio Mount

A patio mount is basically a fixed dish stand which is bolted down. The dish is elevated by a sliding telescopic pipe arrangement, the lower lip of the dish being hinged to the patio frame beneath. Patio mounts are usually found at pub or bookie shop installations where a specific channel, such as CNN, MTV or Sports Channel, is being received since only one channel is required the dish can be permanently fixed for reception from one satellite. BT often instal preset dish systems at bookie shops to receive the betting downlink information service.

Having unpacked my patio mount and dish I had the problem of how to adapt the mount to obtain an azimuth swing so that the dish could be swung from east to west through south, giving access to the Clarke Belt where the
various geosynchronous satellites are parked in orbit. Use of a polar mount, which when set up gives accurate tracking across the Clarke Belt, would have been best, but the impoverished TV-DXer following the set up described here must settle for independent adjustment of the azimuth and elevation.
The patio mount is designed to be bolted down on to a concrete base/flat roof. For this purpose several lugs are welded to the hoop that comprises the patio mount frame. The dish is hinged directly to one side of the frame: at the opposite side of the hoop there's a telescopic steel tube assembly that lifts the dish up in elevation - tightening a single bolt maintains the correct angle. A simple but effective arrangement.

## Obtaining Azimuth Adjustment

I obtained some sturdy industrial casters (try a tool supplier for these) which for fixing purposes have quarter inch threaded studding. The casters can be bolted to the frame to give movement, using appropriate plated nuts and washers. The only problem is that the casters are free to move and rotate on their own. To overcome this difficulty the vertical spindles around which the casters rotate were carefully drilled and tapped through with steel self-tapping screws (one per caster). When fixed as shown in Fig. 1 and the accompanying photos the dish and mount will now rotate circularly, i.e. turn on the spot. Further precision is needed however.

You will probably be able to scrounge an old industrial tidybin lid from your local refuse depot (find under Environmental Health Services of the local council authority). These are flat, circular lids that fit on top of the wheeled bins you find behind shops etc. The type made by Refuse Systems of Bradford is perfectly flat and round (like a tea tray). It has an access hatch with lip - when used in its intended manner a rubber lid fits over this. I acquired a rusty specimen gratis and cut away the small access lip, leaving a large lid with a hole in it. It's best to paint the lid with rust preventer and then a gloss enamel paint.

The idea is that the dish and its now wheeled patio mount sit inside the upturned lid, rotating within the lid in a disciplined manner. Unfortunately the internal diameter of the dustbin lid was found to be greater than the extreme diameter of the casters. To get round this problem sleeved garden hosepipe was fitted around the inside of the lid's outer lip. The hosepipe was made into a loop and joined with a piece of half inch outside diameter alloy tubing -ex-v.h.f. aerial element tubing. The aim is to achieve a friction fit against the side of the casters so that the assembly can be easily rotated but won't move on its own accord. In my case it was necessary to provide additional packing, using thin plastic strip, to obtain the desired degree of frictional pressure on the sides of the casters. Before the wheeled dish and frame are fitted within the


Fig. 1: Fixing the casters to the frame (not to scale). The patio mount casters are fixed at $90^{\circ}$ positions on the flanges provided. Lock the caster wheels (see text) to allow movement in one plane only - parallel with the adjacent frame, as shown here.


Photo 1: The industrial dustbin lid, with access cut-away, treated with an anti-rusting chemical.


Photo 2: The wooden frame, held together with timber connectors, on which the dustbin lid will rest.


Photo 3: The now painted dustbin lid at rest on the wooden frame. Take care not to cut away the access hole lip too close to the main outside lip.
dustbin lid, paint a horizontal white line to serve as the north/south reference. There are several holes in the surface of the dustbin lid. These were rivet holes for the original steel handles. Use two holes to provide anchorage to the ground to prevent movement of the lid and hence disruption of the N/S reference. Leave the other holes to provide drainage.

## Alignment

Make a simple wooden base from creosoted timber, say $2 \mathrm{in} . \times 3 / 4 \mathrm{in}$., holding it in place with "bang-in" timber connectors. Place the lid on the timber frame in a location that gives a clear view of the southern sky between SE and SW. Carefully align the lid with its white horizontal line on a magnetic north/south path. This reference must be


Photo 4: One of the casters bolted to the patio frame hoop. Fix the casters to prevent movement around the vertical spindle. To do this, hold the caster unit in a vice and drill a hole through the side of the caster shroud, then the spindle itself. Fit a self-tapping steel screw of sufficient length to pass through the outside shroud and the spindle, projecting beyond the inner edge of the spindle. The caster will then rotate only parallel to the steel hoop.


Photo 5: The now mobile patio mount sitting inside the dustbin lid. Note the hosepipe packing - this is fixed to the rim of the dustbin lid with Evostick to prevent it moving. The white line is the magnetic north/south reference. The access hole is useful since you can stand in it when adjusting the feedhorn assembly over the top of the dish. Several holes are provided in the dustbin lid to aid drainage and allow two six inch nails to be hammered into the ground to prevent movement of the lid once it has been calibrated with satellite aiming points.
accurate. Avoid using the compass near the lid otherwise the steel will deflect the needle from true. The reason for having the reference line on the lid is so that it can again be aligned when moved elsewhere in the garden and markings for known satellites will remain true.
Make a vertical reference line on the hoop frame, centrally beneath where the dish is hinged, i.e. at the front of the system. When a satellite is located a matching reference line can be painted on the rim of the dustbin lid, with an index or reference number, so that you can always accurately return to the same azimuth.
So we now have a mounting system that provides simple azimuth movement. It cost next to nothing to make (this depends on your sources of scrap metal). Next month we'll give details of the elevation adjustment. If it all sounds a bit


Photo 6: Close-up of the head assembly minus LNB. The support arms with PVC tubing make fixing easy: the protruding screw allows polarisation and focus adjustment. All very simple - you can't go wrong with the equipment I purchased.


Photo 7: Another view of the scalar ring assembly - its adjustment screw can just be seen. This enables the ring assembly to be slid up and down the feed tube. Careful observation of a weak signal on a TV screen while adjusting the ring will show clearly where it peaks.


Photo 8: The Intelsat bird at $27.5^{\circ} \mathrm{W}$ was found within two minutes once the LNB was fixed. Beginner's luck - it took 42 minutes to find the ECS bird at $13^{\circ}$ E! Just to prove that the system works, this photo shows the EBU-Washington news preview feed, not too strong a downlink, being a half transponder on the basic full transponder receiver.
complicated, I should add that I can go from say the Intelsat bird at $27.5^{\circ} \mathrm{W}$ to the Eutelsat bird at $13^{\circ} \mathrm{E}$ in about twenty seconds (plus a walk to the dish!) and later reset to Intelsat accurately without the need for a signal strength meter or a TV screen display.

# TV Fault Finding 

Reports from Mick Dutton, John de Rivaz, B.Sc.(Eng.), D.H. Davies, Hugh MacMullen, Joseph Cieszynski, Roger Burchett and Philip Blundell, Eng. Tech.

## Ferguson TX90 Chassis

This colour portable suffered from intermittent colour. No amount of heat or freezer would induce the fault. We noticed that the colour was always correct when the set was first switched on. It would then go into bars and finally off. From this it seemed likely that the problem was around the colour decoder reference oscillator. We started to change capacitors and when C155 ( 47 pF ) was replaced the colour stayed on. We used a 100 pF component as in later production.
M.D.

## ITT CVC800 Chassis

We'd seen this set several times over the past twelve months. The customer always complained that it failed to start properly. He said it made a screeching noise that built up gradually until the set sprang to life. We were never able to pin this down as the set would always work correctly with the back removed and the test equipment hooked up. Recently however the customer came in and reported that the set had gone dead. On removing the back we found that the 110 V rail was very low and pulsing. It didn't take long for us to discover that the smoothing capacitor C757 $(10 \mu \mathrm{~F})$ was open-circuit.
M.D.

## Philips KT3 Chassis

We were fooled by one of these sets which was tripping. By the time a friend called we'd changed just about everything. "What about the mica washer under the line output transistor?" he said. When removed we found it had a pinhole that produced arcing.
M.D.

## Mitsubishi CT180B

This set had suffered is first breakdown from new (some nine years!) The problem was that the picture was severely reduced in size all round. We found that the h.t. line was at only 85 V instead of 105 V . R909 (220), 10W) was getting very hot and we discovered that the over-voltage protection transistor Q905 (2SC620) had gone short-circuit. A replacement put matters right.
M.D.

## Rank T20 Chassis

This set came in dead. We soon found that the BU208A line output transistor was short-circuit. A replacement was fitted and the usual dry-joints on the line scan plug were resoldered. The set was given a short soak test then returned to the customer.

It came back a few days later with the same complaint. After fitting several BU208As we eventually found that the scan coil connections at the plug on the coils were burnt and making poor contact. Resoldering the wires directly solved the problem.
M.D.

## Amstrad CTV1600

The complaint with this set was field collapse. Supplies were present at the field output transistors but there was no voltage at the base of the driver transistor. On checking back to the LA7800 timebase generator chip we found that there was no 12 V supply to the field section, due to a
dry-joint from the 12 V rail to pin 12. It's worth noting that this i.c. has two supplies - one for the field and one for the line section.
M.D.

## Variac Repair

Having been a wally and burnt out the variac I put it on one side and looked up the prices in a surplus catalogue. After recovering from the shock I thought I'd try to repair the faulty one. It can be seen from the circuit (see Fig. 1) that a burn-out usually occurs when current can flow from the mains to an overload through a relatively short section of the winding. Furthermore, it's the output current that's critical. So a 5A cutout was fitted in place of the more usual terminal block (see Figs. 1 and 2). The prongs were removed and replaced with soldering tags: it was then connected in series with the output cable, which ran to a floating 13A socket.

The variac was repaired by first dismantling the unit completely to reveal the central toroid. The burnt section was wound off and a similar diameter enamelled wire was then selected to wind back in its place. Once the section was rewound it was pressed down so that the turns passing under where the brush would move were flat and level. They were then sanded off with a piece of sandpaper so that the brush could make contact.

If I can make it work after a home repair I'm sure most other Television readers will be able to do the same should they have a similar unhappy accident.
J.deR.

## Monochrome Portable Problems

Dwektronix "Classic 12": No field scan was traced to D506 (BA233) being faulty. It's connected between the emitter and base of the field output transistor. Note that later versions of the Classic 12 have an i.c. field timebase.


Fig. 2: Variac mechanical details.

Monelectric Minimatic 12in. portable: We've had failure of the ITT mains bridge rectifier in several of these sets. Fitting a 3A RS bridge to the heatsink on the right-hand side cures the fault.
Philips TX chassis: Intermittent loss of the sound and vision signals was traced to dry-joints on the tuner's feedthrough capacitors. Resoldering all joints cured the fault.
Binatone Cavalier Model 19496: Loss of sync was traced to D302 being faulty.
Thorn 1590/1 chassis: The l.t. line was low and couldn't be adjusted. The cause was failure of the line output transistor VT26.
D.H.D.

## GEC 14in. CTV (ITT Pico 1A Chassis)

We've had two of these sets with the same problem. If all the tuning buttons are pushed in they will lock in. To release, open the tuning cover and insert a small plastic probe into the plastic plate behind the buttons. Then lift the plate up and down to release.
D.H.D.

## Fidelity ZX2000 Chassis (CTV14)

The switch-mode power supply would trip when the set had warmed up. The cause was C412 ( $1000 \mathrm{pF}, 8 \mathrm{kV}$ ) flashing internally. This capacitor decouples the input to the focus unit.
H. MacM.

## Sanyo CTP3106 (80P Chassis)

This fault took us a time to find because it occurred only when the set was very warm. In this condition there was a considerable increase in the h.t. voltage, from about 115 V to 150 V . The chopper transistor's input coupling capacitor C314 $(47 \mu \mathrm{~F}$ electrolytic) was found to be slightly leaky.
H. MacM.

## ITT CVC1204 Chassis

The switch-mode power supply seemed to be o.k. but the line output stage wasn't doing anything. R744 ( $1 \mathrm{k} \Omega$ ) which feeds the line driver stage often burns up to cause this problem but appeared to be perfect visually. It was nevertheless very nearly open-circuit.
H.MacM.

## Hitachi NP6C Chassis

This was a difficult one because the fault occurred only with the back on! The picture would suddenly shrink to about 12in. but remain quite linear and in focus. Eventually we found that TR904 in the regulator circuit was intermittently faulty. The h.t. would then fall to about 95V. H.MacM.

## Philips K30 Chassis

No video and a very dark raster can be caused by a number of things with these sets, but two faults of particular interest came our way recently.

In the first case there was no luminance output from the TDA2560) chip (earlier two-chip decoder). After some time I tried another decoder panel and proved that the fault was on the mother board. Checks around the flyback blanking transistor T1535, the contrast controls, etc. revealed nothing but I then discovered that pulses were present at the cathode of D1422, which links the line output transformer derived beam-limiting potential to the contrast circuit. This provided the clue since there should be a steady voltage at this point. Inspection of the 68 nF smoothing capacitor C1565 revealed a hairline crack at one
end, and when a replacement was fitted a normal picture appeared.
The second set had the same symptoms and initial checks such as unplugging the teletext interface and checking the d.c. voltages showed us only that the tube's cathodes were at 150 V , so that the tube was cut off. A scope revealed that colour-difference and luminance signals were present at the decoder panel, but tests around the RGB output panel led us nowhere. At one point we checked the -20 V bias supply from the line output stage and found that it was slightly high. Not too surprising perhaps as the supply is unregulated, but we nevertheless found that the relevant reservoir capacitor ( $\mathrm{C} 1586,100 \mu \mathrm{~F}$ ) was open-circuit. We discovered this by scoping the -20 V line which turned out to have line-frequency pulses on it. Interesting that when measured with an Avo 8 the voltage was found to be high, something that shouldn't happen when the supply decoupling is open-circuit. We can only surmise that the presence of line pulses with the meter switched to the d.c. range led it to produce an incorrect reading.

Other causes of no picture with these sets are loss of line pulses due to dry-joints, failure of the TDA2560 or TDA3560/1 chip (depending on decoder type), failure of the flyback blanking transistor (T1535) and, on teletext versions, failure of either fuse on the teletext power supply mounted at the base of the cabinet.
J.C.

## Panasonic TC2201

A sad story this. The elderly owner complained that there was intermittent line collapse. No fault could be found during a number of visits and eventually the set failed. Fusible resistor R525 ( $68 \Omega, 0.5 \mathrm{~W}$ ) in the feed to the line driver stage had gone open-circuit. A couple of ordinary $33 \Omega$ resistors in series were fitted temporarily and everything seemed to be all right, but the set failed again a short time after fitting the correct type of resistor. This time the BU208A line output transistor had failed. The set seemed to work normally after replacing this, though the calls complaining about intermittent line collapse were becoming more regular. Finally I saw what the problem was: the set was tripping due to the h.t. rising. I could see the h.t. breathing as I watched it.

On Panasonic's advice I replaced the two zener diodes D809 and D819 which. incidentally, are 5 V and 6 V respectively, not both 6 V as stated in Service Bureau last September (page 773). To be on the safe side I also replaced D815 and the set-h.t. control R813 ( $1 \mathrm{k} \Omega$ ) which has been known to give trouble.

Unfortunately the focus control had suffered too many blows and was varying intermittently. It's no longer available from Panasonic, so a Thorn 8500 type was pressed into service.

Reverting to the Service Bureau item just mentioned, D815 and R813 should be added to the list of items to be replaced.
R.B.

## Decca 70 Series Chassis

I've had quite a few colour portables in recently fitted with this chassis and with the complaint no colour. The cause of the trouble is often a faulty chroma delay line (DL700). Due perhaps to the sets being dropped? P.B.

## ITT 80-90 ${ }^{\circ}$ Chassis (Power board CVC820)

If the line output stage is drawing excessive current, pulling down the 120 V h.t. supply, before suspecting the shift or line output transformer try disconnecting pin 5 of
the TDA1170S field timebase chip in case it's shortcircuit.

If the set changes channel intermittently by itself, try changing the focus spark gap.
P.B.

## ITT CVC30 Chassis

I seem to see a lot of ITT sets: this day all the calls were to the same range. Some of the faults were stock ones, some not.
(1) Dead with the power supply whistling and no 160 V h.t. supply. The h.t. rectifier D19 was open-circuit. Someone had fitted an SKE type.
(2) Dead, power supply tripping. The BU208 line output transistor was short-circuit due to a burnt scan coil plug.
(3) Lack of height due to dry-joints on the field module earth.
(4) Intermittent dark picture due to dry-joints on the EW modulator transformer.
(5) Dead set, no 160 V supply to the line output transistor due to a dry-joint on the tag to mounting bolt.
(6) Remote control not working when out of set. PCB mounted coil in handset broken off.
(7) Blank raster. R28 (820 ) open-circuit. This resistor feeds line pulses to the colour decoder panel.
(8) Tripping - stopped when the tripler was disconnected but the fault was still there with a new tripler fitted. C61 on the earthy side of the line output transformer's e.h.t. overwinding was short-circuit.

Didn't need the loan set that day!
whether to close it down completely. Elimination of manufacturing losses would enable Fidelity to operate profitably.

Meanwhile Hinari has decided to establish a new plant at Cumbernauld, Scotland for the production of brown goods. Initial products will be hi-fi and compact disc systems.

## s-VHS

JVC has announced that the S-VHS specification for Europe has now been finalised. Initial information on S-VHS was given in this magazine last July.

The main change with S-VHS, which requires special tape, is that the f.m. carrier deviation is $5 \cdot 4-7 \mathrm{MHz}$ instead of $3 \cdot 8-4 \cdot 8 \mathrm{MHz}$, giving a horizontal resolution of over 400 lines. The chrominance carrier frequency remains the same at 627 kHz while the white and dark clip are 210 and 70 per cent respectively. Audio is standard VHS linear or VHS hi-fi. Tapes will be available in a variety of lengths, designated SE180, SE120 etc. - there's so far no provision for a four-hour tape. S-VHS-C tape is the standard thirty minute length (SP mode). European S-VHS equipment is being designed to work with PAL, Secam and other signals including MAC. No launch details have been released.

An International Electrotechnical Commission (IEC) proposal for scart connectors to be used with the new S-VHS system is under consideration. The IEC suggests splitting the separate chroma signal, used with S-VHS to avoid cross-colour effects, into RGB components to feed to the appropriate scart socket pins. This would involve additional circuitry but would have the advantage of compatibility with TV sets equipped with a scart socket.

## PRODUCTS GO DIGITAL

Digital seems to be the flavour of the month as Japanese VCR manufacturers add digital machines to their ranges. Models from Panasonic, Toshiba and Sharp have already been mentioned in these pages. Toshiba's latest model, the DV-90W, has digital still and slow plus HQ Pro, which employs all four of the HQ features and additional filtering. It also has an elapsed time counter which uses the control pulses for time calculation. The price is around $£ 480$. NEC's latest VCRs, Models DX-1000K and DX-3000K, incorporate a digital video noise reduction system and digital still, slow and picture memory (the off-tape or off-air picture can be frozen). Other features include twin-speed strobe, fast search and HQ circuitry. The DX-1000K has a suggested price of around $£ 450$ and the DX- 3000 K , with hi-fi stereo sound and two-speed operation, a suggested price of around $£ 700$.

A digital TV receiver using ITT's chip set has been
introduced by Telefunken. It has a 29in. "Super-Planar" flat-face tube with a new type of gun assembly and a stereo sound capability of 40 W music power per channel with four front-facing speakers. The colour decoder is of the multistandard type, handling PAL, Secam and NTSC signals, and the coverage includes v.h.f.

Panasonic has introduced a digital mixer which can mix or wipe any two video signals whether synchronised or not. Special effects can be created by a frame synchroniser. The WJX10 sells at around $£ 1,200$ ).

## TV SOUND TUNER

To overcome the problem of poor sound reproduction from the average TV receiver Radio and TV Components (Acton) Ltd. has introduced an independent TV tuner which can be directly connected to a hi-fi system. The unit is mains operated and has full u.h.f. coverage with five preselected tuning controls. It can also be used in conjunction with a VCR. The basic model costs $£ 29 \cdot 50$. A version with a built-in stereo headphone amplifier for the hard of hearing costs $£ 35.90$. The tuners are available by mail order (add $£ 2.50$ post and packing) from the company's Acton branch at 21 High Street, London W3 6 NG . The company has branches in Acton and the Edgware Road.

## LOW-NOISE HYBRID AMPLIFIERS

The new OM2000) series of Mullard hybrid amplifiers offers improved performance with MATV, CATV etc. systems noise figures are about 25 per cent less than with the standard range. There are five devices with type numbers OM2045, OM2050, OM2060, OM2061 and OM2070. Performance ranges from a gain of 12 dB with noise figure of $3 \cdot 6 \mathrm{~dB}$ for the OM2045 to a gain of 28 dB with a noise figure of 4.8 dB for the OM 2070 . The amplifiers are wideband devices covering $40-860 \mathrm{MHz}$ with input and output at $75 \Omega$ and require a 12 V supply. They use thin-film technology and the latest Mullard u.h.f./s.h.f. transistor type BFR92.

## IERE RECORDING CONFERENCE

Sessions on laser recording techniques and media will be featured at this year's International Conference on Video, Audio and Data Recording, which is being held at the University of York from March 21-24. The conference is the seventh in the biennial series organised by the Institution of Electronic and Radio Engineers. It will be preceded by a tutorial day wholly devoted to optical recording. Two papers on erasable optical storage, from Philips and Sharp, will be included in the first session of the main conference. For further information and registration forms apply to: The IERE Conference Secretariat, Savoy Hill House, Savoy Hill, London WC2R 0JD (telephone 01-240 1871).

## NEW ME TAPE PROCESS

Thorn-EMI has developed a new process for producing metal evaporated (ME) tape, which is used for Video 8 and digital audio cassettes. The process involves enclosing the tape coating plant in a vacuum chamber. A crucible of metal is then heated to boiling point with a high-power electron beam. Finally a wide roll of polyester is wound over the crucible so that the metal atoms condense on the backing to produce an 0.15 micron recording layer. Special shutters ensure that only atoms arriving at an oblique angle
are deposited on the backing. According to Thorn-EMI this results in the columnar crystals leaning against and supporting each other instead of standing upright on the polyester. The magnetic resolution is claimed to be better than 0.3 microns.

## FLAT TRADING

Disappointing half year results announced by Dixons have underlined the fact that the radio/video/TV trade has been going through a flat patch, with a poor Christmas. Dixons ordered ten per cent more stocks for the Christmas period but sold twelve per cent less. The present aim is to clear excess stocks by price cutting.

## VIDEOPHONE FORMAT AGREEMENT

Agreement of a standard for videotelephones has been reached by Japanese manufacturers and has been given preliminary approval by the Telegraph and Telephone Technology Committee. The standard relates to the transmission of still pictures over conventional telephone lines. Transmission of a picture takes five-six seconds, during which time conversation is not possible. Mitsubishi, Matsushita, NEC and Sony plan to start selling still picture phones this spring. There had been disagreement earlier between Mitsubishi and Sony, both of whom have been selling videophones in the USA. Mitsubishi's system was quicker and had a slightly larger screen, but sold for $\$ 1.000$ compared to about $\$ 375$ for the Sony equipment. The systems basically use fax principles with a camera and microchips for image processing.

## NEW APPROACH TO 3-D VIDEO

A small, London-based technology company, Aspex, has developed a new approach to 3-D applicable to all forms of recorded visual images including TV, video and film. The system uses a special lens - there's no need for two separate images - and is said to improve the sharpness and colour saturation of images. When special glasses are worn a depth effect is created.

## telecom still picture transmitter

Canon has developed a portable transmitter, type RT611, for use with its still video communications system. The transmitter enables images held on an SV floppy disc to be easily transmitted via an ordinary telephone to anywhere in the world. It can be connected to the line directly by means of adaptor LC-RT or coupled to the telephone mouthpiece by means of acoustic coupler AS-RT. The RT611 has a built-in 1.5 in. screen for display of the image. At the receiver end a conventional wire phototransceiver can be used for monochrome or a Canon SV transceiver type RT971 for colour or monochrome. Colour transmission takes only three minutes.

The transmitter was exhibited at the recent Geneva Telecom Show alongside two units currently under development, the Canon image processor and a video input adaptor for use with the Canon digital colour laser copier. The latter was launched in the UK last year. The image processor will process video images in a variety of ways capabilities include manipulation of form, colour and image size, and combining multiple images to form a composite image. The unit is computer controlled and incorporates a frame store: to alter the image as required a pressure pen is used with a colour monitor.

# Long-distance Television 

## Roger Bunney

Apart from an uplift from Geminids/Ursids meteor shower activity December is traditionally a quiet month. This time it was somewhat different. There were no fewer than three tropospheric openings that produced quite excellent reception, particularly in central/southern England, during the period.

The first spell occurred on December 6th, when a prevailing high-pressure system produced high-level Band III/u.h.f. signals from near/central Europe. Typically Belgium, Holland, France, West/East Germany and Denmark were received in the Midlands.

The second spell occurred around the 14th, with a virtual repeat of the conditions on the 6th though Band III ducting was more noticeable - many enthusiasts received CST (Czechoslovakia) ch. R10 (PIzen) for the first time. The opening continued through the 15th, with reception extending as far as Poland. Another first for several vigilant DXers was TVP (Poland) ch. R38 Wrocklaw. One comment had it that this transmitter came in "like a local". Reception of CST ch. R35 was also widely reported.

The third and perhaps most dramatic opening occurred on the 23 rd, though it was trailed and tailed on the 22 nd and 24th. A fast-moving lift produced rapidly changing and selective reception, with ducting. Towards the latter part of the event signals were received from Scandinavia. Several logs received resemble a West European transmitter list, covering from RTE (Ireland) in the west to Denmark in the east and NRK (Norway) to the north. France was well received, with David Moller in Birmingham logging TV5 on chs. E29 and E35. Several DXers had their first sighting of the new NOS-3 (Holland) ch. E34 Roermond transmitter on test pattern. Many West German Band III and u.h.f. stations were logged, and as with the earlier openings Band III was most rewarding, with TVP-1 ch. R8, SR (Sweden) ch. E8 and u.h.f., NRK chs. E5, 8 and 9 (but no reports of the new u.h.f. relays!), RTL (Luxembourg) ch. E7 and, for those near the east coast, Dutch ATV amateurs (PE1HLR, PE1DWA) in the 435 MHz band. An interesting reception for three DXers was the ch. E2 100W BRT (Belgium) relay in Antwerp, with vertical polarisation - even Simon Hamer in distant North Wales logged this one!

Three doses of tropospheric reception during December constituted a good Christmas present for many TV-DXers. It's unfortunate that 435 MHz ATV activity seems to be on the decline. Though they do look, few DXers now report having seen any ATV transmissions during good conditions. Perhaps there's been a mass migration to 1.3 GHz f.m.!

There was some Sporadic E reception during the month. The collated $\log$ is as follows:

[^1]23/12/87 TVE E2.
25/12/87 TVE E2, 3; NRK E4.
26/12/87 TVE E3.
27/12/87 RAI 1A; NRK E3.

A very slight tropospheric lift was noted on the 27 th, with mainly signals from TDF (France).

Auroral activity was very quiet. Iain Menzies (Aberdeen) noted slight disturbances on the 12th, 14th and 19th.

The tropospheric activity turned interest away from MS reception - the Geminids and Ursids showers seem to have produced minimal activity this year. The January Quadrantids around the 4th produced an increase in the normal diurnal activity, with Band I favoured - no reports of Band III reception at all.

An excellent month for December then, ending the year with a flourish!

My thanks to the following for their reception reports: Iain Menzies (Aberdeen), Simon Hamer (Powys), David Oliver (Birmingham), Gareth Foster (Twickenham), Cyril Willis (Norfolk) and Roger Fussell (Torpoint).

George Gaskin (Gibraltar) reports that TVE is now operating for 24 hours a day at weekends and that private stations will be starting up over the next two years, also a third channel in the Andalusian region. So we should be noting more Spanish reception. For optimists, GBC-TV (Gibraltar) has started its "infotel" service, a continuous series of advertisements outside broadcast hours, generally on a 24 -hour basis.

During the past month I've been assembling a flexible TVRO system using a 90 cm dish with patio mount and 11 GHz satellite package. The results are chronicled elsewhere in this issue. The aim was to gain experience in this new field (following earlier experiements at 4 GHz and with the 860 MHz ATS satellite) and to encourage others. It can be an expensive move to make, so I opted for the cheapest solution possible which has meant operational limitations. On the day this was written, January 7th, I noted a new downlink on the ECS bird at $10^{\circ} \mathrm{E}$. The 11.65 GHz (horizontal) signal consisted of colour bars with the identification E8T-5-MI and conversation in Italian. There are signals apart from Super Channel and the domestic/ cable downlinks about.

## News Items

UK: New scope for TV-DXing in the UK is in prospect with the efforts being made to find space for fifth and sixth networks. Many new transmitters could be accommodated in chs. 35-38, the problem being that parts of these channels are at present used for airport radar and radio astronomy. Another possibility being considered is distribution at 2.5 GHz (see article in the November issue). For this latter application North East Satellite Systems has already made prototype receiver-converters with six inch dishes, aiming for a price at around $£ 50$. The signals would be down converted to u.h.f./a.m. at the head. Using current technology, systems could be in operation within eighteen months. A microwave band that's likely to be allocated to truly local terrestrial TV within the next five years is at 29 GHz .
Devices called videosenders are currently available at various glossy high street hi-fi stores. They are illegal to use but not to buy! Their purpose is to enable the user to transmit the output from his VCR around the house - and it seems around the immediate neighbourhood as well. Garry Smith (Derby) recently tested one and found that it
produced excellent quality radiation even without fitting an aerial. Ranges claimed are up to 165 ft , at around ch. 21 . Gareth Foster has taken the use of these devices to the Advertising Standards Authority, since they are being advertised in video magazines with in some cases no warning about the illegality of their use. Interference to IBA ch. 21 transmissions has already been investigated. Belgium: A new TV service in Flanders, Vlaamse Televisie Maatschappij, is due to start this autumn. It will carry advertisements. Operators have yet to be appointed. Australia: The first Aboriginal TV service, Imparja Television, has been brought into operation by the Central Australian Aboriginal Media Association of Alice Springs. The transmissions are uplinked to the Aussat satellite, picked up on downlink by a number of ground stations and then retransmitted locally. Local stations can opt out of the network Imparja programme.

The Minister for Communications has announced a timetable for the clearance of TV from Band II (chs. 3, 4, 5 and 5 a). Services will be moved to other frequencies, including u.h.f. Most stations are to be moved by 1993 though a few relays will continue in operation until 1/1/96. The aim is to allow more rapid development of the Band II f.m. radio services. Australian readers can obtain the "Television Station Draft Clearance Timetable", media release no. $98 / 86$, from the Department of Communications (062 $64 \quad 3235$ )
West Germany: The opening up of u.h.f. channels E61-68 for TV use is progressing well and it's hoped that the new spectrum will be in use by private TV starting this autumn. Low powers will be used initially, with higher power stations later. There's more DXing potential with transmitters in the Schleswig-Holstein area transmitting the SAT-1

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programme (Garding ch. E25 at 6kW e.r.p., Schleswig ch. E42 at 330W and Eckernforde ch. E37 at 120W) and RTL Plus (Garding ch. E58 at 6 kW , Suderlugum ch. E56 at 5 kW , Schleswig ch. E52 at 500W and Eckernforde ch. E60 at 120 W ).
Denmark: Two additional TV2 stations are likely to come on-air this summer (TV2 officially starts in October). These are TV-Fyn and either TV/nord (Aarhus transmitter) or a unit in Copenhagen.
Radio Amateurs: Dutch amateurs have been allocated the $50-50.45 \mathrm{MHz}$ spectrum for c.w. (morse) at up to 30 W output from March 1st to January 1st 1994, with an annual review by the authorities to consider any interference problems - we understand that many cable systems now use ch. E2.
Computer interference: Interference from the BBC computer can be reduced significantly. An article in the December issue of the RSGB magazine Radio Communication contains a thorough discussion of the problem and practical suppression details that allow reception at 50 MHz with the receiver some ten feet from the computer!
Satellite TV: As mentioned in Teletopics last month the West German TV-Sat 1 DBS satellite has been written off. The loss also affects the start of the French DBS service and, we understand, later stages of the Aussat programme.

The UK film channel Premiere is now using SAVE-SAT scrambling.

## Signal Strength Meter Postscript

I reviewed the Planet SSMU signal strength meter, a budget priced instrument distributed by HRS Components, in the March 1987 issue. The review seems to have had an effect since the two shortcomings noted have been dealt with in the new SSMU-2 version. Aerial Techniques have sent me an instruction folder which shows that an increased audio level is now available via a two-position switch (for normal or boost level) and that a 3.5 mm jack socket has been fitted for external headphone use outdoors - inserting a 3.5 mm plug cuts out the internal sound via the case-mounted transducer.

## Studies on Additional TV Services

A press release from the DTI, dated December 17th, brings good news. The current studies into the feasibility of
additional TV networks for the UK are to be extended to include Bands I and III as well as u.h.f. and MMDS at s.h.f. ( $2 \cdot 5 \mathrm{GHz}$ ). The DTI seems to prefer the term MVDS (multipoint video distribution system) to MMDS (multichannel microwave distribution service). At v.h.f. the study will consider the possible effect on recent mobile radio allocations.

## From our Correspondents...

Keith Watkins (Redruth) has written to us on some interference problems. A friend lent him an RTTY/CW decoder unit which wipes out Band I, while his Christmas present, a Philips CD160 compact disc player, similarly removes any chance of Band I reception. Has anyone any solutions for this latter problem?

Fred Robins (Stubbington) spent a period in Japan recently. During his visit he took a series of excellent off-screen photographs of local TV test patterns etc. We'll be featuring some of these over the next few months. Thanks Fred.

Jean Louis Dubler, who has written to us previously from South Korea, has now moved to Montreux, Switzerland. A recent letter describes local TV conditions there. He has four Swiss and four French channels, two of which (Canal Plus and Telecine) are scrambled. The situation is about to change, with Canal Plus taking over the Swiss Telecine transmitters. The two-year-old Telecine has always operated at a loss, but Canal Plus has sufficient subscribers to make a profit. Canal Plus will have to use a different scrambling system in Switzerland since pirate Italian decoders are available there cheaply. It has opted to use the Telecine system. The situation is further complicated since Canal Plus uses yet a third system in the Swiss/French border area. Canal Plus operations are eventually to be extended to Belgium and Morocco.

A pirate station near Geneva transmits on ch. E52 using the SECAM-L system and intends to introduce repeaters, assisted by the NRJ f.m. radio network. There's been a proliferation of pirate transmitters in the French/Swiss border area, some like Radio Thollon at $93 \mathrm{MHz}, 4 \mathrm{~kW}$ operating with high output powers. Many stations on the French side of Lake Geneva aim at a Swiss audience.

The Montreux cable network has fifteen channels, including the English-language Sky and Super services and the French LA5 and M6 services.

## A Professional Institution for TV Technicians

lan Channing

There have been professional bodies in the consumer electronics industry since the early days of broadcasting. In the 1920s the Institute of Wireless Technology was formed: it eventually became the Institution of Electronic and Radio Engineers. The Institution of Practical Radio Engineers was founded in the 1930s: this became the Incorporated Practitioners in Radio and Electronics. The Guild of Radio Service Engineers appeared in the 1940s, only to disappear in the 1950s. The Society of Electronic and Radio Technicians (SERT) was founded in 1965, to look after the interests of what are now known as Engineering Technicians and Technician Engineers. In every case the aim of founding such bodies was to establish a standard which employers would recognise and use. It would keep out the "cowboys". The problems are that (a) there will always be cowboys around and (b) the industry
needs a means of identifying competent staff.
Over the past twenty years the pattern of qualifications has changed. The majority of technical staff in the servicing field now qualify through the City and Guilds Course 224, Electronic Servicing, which however does not meet the current requirements of the Engineering Council for Technician registration. Realising this, SERT sought a means of providing a professional service for servicing staff who, though qualified, were not eligible to join the Society.

The opportunity to do this occurred in 1982 when SERT was asked to take over the Incorporated Practitioners in Radio and Electronics (IPRE). This body had been in existence since 1935 but had been going through a period of considerable decline. When the existing members of the old IPRE were absorbed into SERT a new division was set up, using the same title - the IPRE Division. The Society's

IPRE Division provides a complete professional service for qualified staff engaged in the maintenance, test and installation field in all branches of electronics. It has an autonomous Board Management which runs its own affairs and has representation on the main Council of SERT.

There are two corporate grades within the Division Member and Associate Member. Members of both grades are entitled to use the appropriate designatory letters MIPRE or AMIPRE. There is also a Student grade for those still receiving technical education.

All the Division's members enjoy the same learned society benefits as members of SERT. These include the monthly journal Electronic Technology, which contains feature articles, industry news and new product information across the whole range of electronic engineering. The Division organises special one-day seminars on matters of current technical interest, such as compact discs and satellite broadcasting. All IPRE members are entitled to attend these at a reduced members' rate. They are also entitled to attend seminars and residential conferences organised by the Society of Electronic and Radio Technicians.

IPRE members are able to participate in SERT Local Section activities, and most section committees include at least one IPRE member. These activities include technical lectures, visits and social events.

The main qualification for membership of the IPRE Division is the Part II Certificate of Course 224, Electronic Servicing, but certain service and company qualifications are accepted on an individual assessment basis. The minimum age of admission to the grade of Member is 26 : applicants must be exercising some degree of responsibility such as being a senior engineer or service manager. Associate Member applicants must be at least 20 years of age and have had one year's appropriate experience.

Membership at present costs $£ 18$ a year for Members and $£ 16$ a year for Associate Members. There’s a $£ 5$ entrance fee for both grades. Student members pay $£ 7$ a year and there is no entrance fee.

The aim is to maintain standards and in so doing enhance the status of appropriately qualified personnel. Membership enquiries are welcome. Full details and application forms are available from the Secretary, IPRE, 57-61 Newington Causeway, London SE1 6BL.

## Fast-shutter Video Cameras

Eugene Trundle

For conventional applications the pick-up device used in a video camera, whether of the broadcast or consumer type, integrates the received image over an entire field period. During this period each picture element (pixel) at the rear of the pick-up device's faceplate charges or discharges depending on whether the image sensor is a solid-state type or a vidicon-type tube. With a vidicon the scanning electron beam charges the photosensitive surface once per field: between scans, the surface discharges depending on the intensity of the light falling upon it. With a solid-state image sensor the photosensitive surface charges, the signal being read out by charge transfer at field rate. Whichever way it's done, the important thing is that there's a storage effect during each TV field.

As with conventional still photography, this long exposure time gives good sensitivity. Unfortunately however it means that fast-moving objects are blurred. If anything in the picture moves appreciably during the 20 msec field period it will be reproduced as a blur, no matter how good the still-frame arrangements employed by a VCR used to play back the picture. The problem is that the video signal at each pixel represents the integration of all that's visible during the whole field period, not just the brief moment when a pixel is being scanned or read out.

With a vidicon type tube little can be done about this. Either a very special target layer would have to be used or a fast-scan system with some form of external field storage. Neither is practical for an inexpensive camera with the requirement to revert at will to conventional image sensing. With a CCD (charge-coupled device) type of solid-state image sensor however the scanning and storage functions are easy to separate. This opens the way to the use of fast-shutter techniques which give clear reproduction of fast-moving objects.

The idea is that each sensor pixel is blinkered during most of the duration of the field period, taking a very brief "peep" at the scene at 20 msec intervals. Again as with conventional still photography there's a penalty to be paid:
light sensitivity is inversely proportional to shutter speed, so that a camera operated in the fast mode will produce good pictures only when the light conditions are good.

## CCD Operating Principles

Behind the faceplate of a CCD image sensor there's an array of capacitive photodiodes arranged in rows and columns. These correspond with the lines and pixels that make up the TV image. With suitable biasing each photodiode acquires a charge that corresponds to the light level it sees. The imaging surface of the CCD consists of hundreds of thousands of mutually isolated photodiodes. The output from each photodiode is connected to a MOSFET transistor that acts as a switch - see Fig. 1. When a pulse is applied to the gates of these transistors the


Fig. 1: Photodiode charge transfer. The diodes on the left represent one column of image sensors: the "progressive bucket" effect of the transfer charge voltage applied to each cell of the register on the right is represented by the depth lines on their sides.
charges on the photodiodes are transferred to a shift register. Unlike a digital shift register, the type used in this application can handle an analogue signal that consists of charges, or "packets" of electrons. This type of shift register is commonly referred to as a bucket-brigade device (BBD).

Shifting the charge packets along the register is achieved by sequentially altering the potentials applied to the BBD's cells. The electron packets have a tendency to fall into an adjacent "potential well": by creating successively deeper depletion layers in adjacent cells the electron packets can be stepped along the shift register/BBD by using clock pulses in a four-phase sequence.
As shown in Fig. 2, each column of photodiodes has an associated, separate vertical shift register. During each field blanking period a transfer pulse is applied to the gate of each FET. As a result the charges developed by the photodiodes are transferred to the associated vertical shift registers. All the FETs are switched on at the same time, so that once per TV field a complete set of pixel charges is stored.

On the first change of V-clock pulse the charges in all the vertical shift registers move up one. At the top there's a horizontal shift register which thus receives the first line of the picture. This is another BBD, whose contents are now rapidly transferred leftwards by a second and much faster four-phase clock pulse system. The charge packets fall off at the end of this shift register as it were, forming a sequence of pulses of varying height - the analogue video signal. The clock pulses have to be filtered out before the signal can be used.

During the line blanking interval the vertical registers are again pulsed, so that successive complete TV lines are fed into the horizontal shift register. These charges are clocked leftwards along the horizontal shift register during the following line scan period. We thus get at the output a serial information stream that corresponds to the target output from a conventional vidicon tube. At the end of a field the charges from all the photodiodes have been read out, the vertical and horizontal registers are empty, and the
whole sequence is repeated. The CCD clock and drive pulses are provided by a timing/divider chip which is governed by the camera's master subcarrier and sync generator (SSG) section. This is in turn controlled by a precision crystal.

## Timing

Each photodiode or pixel sensor is briefly addressed once per field. Between times it sits there building up a charge depending on the light input - see Fig. 3. The pulse train at the top of this diagram represents the field blanking intervals. At time $t 1$ we are approaching the end of a field period and charges will have been built up on the photodiodes which have for some time been isolated from the vertical shift registers. The video information from the previous field - A - has been moving along the vertical shift registers as shown in the lower half of the diagram. At time t2 the transfer pulse occurs, during the field blanking period. The next field, $B$, is then fed into the vertical shift registers, ready to be clocked through. The photodiodes are now discharged and start to charge once more to produce the next field C .

## Fast-shutter Mode

Fig. 4 shows the sequence of events when the CCD control chip is switched to the fast-shutter mode. Again at time tl we are towards the end of one field period and each photodiode has had some time to charge. This stored image will contain blur, and must therefore be discarded. At about line 623 a transfer pulse $t 2$ dumps the charges into the vertical shift registers. Soon afterwards a high-speed charge-shifting pulse train is applied to the shift registers to flush them clean - see Fig. 5. The effect of this is not seen it occurs during the field blanking period, when the video is muted.

Meanwhile the photodiodes have again been charging. They are allowed to do so for 19 TV lines (nos. 623 to 17), as Fig. 5 shows. On line 19, at time t 3 in Fig. 4, a second


Fig. 2: Representation of a complete CCD image sensor, simplified to show 64 sensors in an eight-by-eight matrix. Typical practical arrays for consumer cameras would have about 250,000 elements arranged in a $579 \times 422$ matrix. The switched charges progress upwards along the vertical shift registers then along the horizontal shift register, under the control of fourphase clock pulses.


Fig. 3: Photodiode charging and transfer switching at the normal $(20 \mathrm{msec})$ shutter speed. Three successive fields, identified as $A, B$ and $C$, are shown.


Fig 4: Photodiode charging (two-cycle) and transfer switching in the high-speed shutter mode. Between t2 and t3 a fast discard clocking pulse train sweeps all the C information out of the vertical shift registers.


Fig. 5: Time-related waveforms for the high-speed shutter mode, (a) line sync pulses, (b) field sync pulse, (c) composite blanking, (d-g) three-level vertical drive pulses in the image sensor section - the highest levels V1 and V3 trigger photodiode charge transfer. The 805 kHz pulse trains in the waveforms on the left rapidly empty the vertical shift registers during the first half of the field blanking period.
transfer pulse is applied to once more fill the vertical shift registers. This time the information contains no blur, since the "shutter" has been "opened" for only a 19 -line period. Nineteen lines is 1.216 msec or $1 / 822 \mathrm{sec}$. The brightness information obtained during this short period is stepped along the registers in the normal way to form the video output signal. This is continuous because of the storage effect introduced by the cells in the BBD shift registers, but at a lower level, as Fig. 4 shows.

There are other methods of carrying out fast-shutter operation. The one described, used by Panasonic, doesn't require special facilities on the sensor array itself. A range of shutter speeds can be provided to trade off sensitivity against image blurring. You can't however see how well you've done until you get home and use the VCR to freeze the image, so that any shutter speed control is best done automatically with reference to the available light. No doubt the next generation of fast-shutter cameras will have automatic movement speed detectors and light meters
hooked to a microcomputer to govern the exposure time. With auto-focus, auto-iris and auto-white balance they may, by 1990, have disappeared up their own exhaust pipes: we can then all go back to box Brownies.

## Use

The fast-shutter facility should be used only when it's known that freeze-frame reproduction with a suitable VCR will be required. This avoids not only the penalty of a twenty-fold decrease in light sensitivity but also the loss of some "smoothness" in the picture when it's viewed in the normal playback mode.

Fast-shutter operation is really successful only in sunlight: artificial lighting from an a.c. source (i.e. the domestic mains supply) gives rise to a heavy flicker effect. This is most noticeable with fluorescent lighting which has short-persistence phosphor. Fluorescent lighting makes camera colour balancing difficult anyway.

| 15/80 H | 3.30 | 2SA940 | 1.32 | 2SC535 | 0.79 | AF180 | 0.55 | BA656 | 1.00 | ${ }^{\text {BC5 } 50 C}$ | 0.14 | BDX63A | 196 | BFY52 | 027 | BYX71-350 | 0.72 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1585R | 3.30 | 2SA940-2 | 214 | 2SC536 | 0.45 | AF181 | 0.53 | 8A7100 | 11.35 | ${ }^{\text {BC63 }}$ | 0.36 | BDY20 | 121 | BfY79 | 0.49 | BYX94 | 0.16 |
| 16039 | 0.79 | 2 SA950 | 0.72 | 2SC537 | 0.54 | AF188 | 0.53 | Babsia | 28.98 | BC636 | 028 | BDY81 | 1.05 | BFY90 | 0.61 | BYY5 | 120 |
| 16181 | 1.04 | 2SA951 | 1.75 | 2SC605L | 1.16 | AF239 | 0.43 | ba843 | 3.96 | BC637 | 024 | BFI15 | 0.40 | BLY49 | 220 | ВZY93С30 | 1.86 |
| 16182 | 1.04 | 2SA966-Y | 1.16 | 2SC620 | 0.95 | AF279 | 0.88 | BAB54 | 5.76 | ${ }^{\text {BC639 }}$ | 020 | ${ }^{\text {BFIL17 }}$ | 0.66 | BR100 | 029 | BZY88 RANGE | 0.10 |
| 16334 | 0.98 | 2SA999 | 1.36 | 2SC643A | 1.54 | AL113 | 1.36 | BAV18 | 021 | ${ }^{\text {BC640 }}$ | 024 | ${ }^{\mathrm{BF} 5118}$ | 0.67 | BR101 | 0.65 | BZX61 RANGE | 0.18 |
| 16335 | 0.94 | 2SB774 | 1.15 | 2SC668 | 0.67 | ANT15 | 3.98 | BAV19 | 0.11 | ${ }^{\text {BC879 }}$ | 0.49 | ${ }_{\text {BFI2 }}{ }^{\text {FF/ }}$ | 025 | BR103 | 0.55 | BZX79 RANGE | 0.10 |
| 16446 | 0.98 | ${ }^{2 S B 185}$ | 1.13 | ${ }^{2 S C 681}$ | 4.40 | AN155 | 1.89 | BAV21 | 0.12 | ${ }_{\text {BC }}$ | 0.18 | ${ }_{\text {BF127 }}$ | ${ }_{0}^{0.13}$ | ${ }^{\text {BR } 303}$ | 1.15 | ${ }^{\text {C106D }}$ | 0.46 |
| 16600 | 1.38 | 2SB375 | 3.87 | 2SC682 | 1.88 | AN206 | 258 | BAW62 | 0.11 | ${ }_{\text {BCY70 }}$ | 0.30 | ${ }_{\text {BF1 }}{ }^{\text {b }}$ | 029 | BRC116 | 0.67 | C106M | 0.76 |
| 16802 | 127 | 2SB400 | 0.40 | ${ }^{2 S C 684}$ | 1.05 | AN208 | 3.55 | BAX12 | 0.48 | ${ }_{\text {BCY }} 1$ | 021 | ${ }_{\text {BFF }} 153$ | 0.58 | BRC300 | 201 | ${ }^{\text {C12 }} 12$ | 0.58 |
| 17052 | 5.61 | 2SB405 | 1.03 | ${ }^{2 S C 693}$ | 0.63 | AN210 | 228 | bax 13 | 0.11 | BCY72 | 0.20 | BF554 | 026 | ${ }^{\text {BRC5296 }}$ | 0.7 | CA3046 | 1.55 |
| 17053 | 5.61 | 2SB449B | 6.98 | 2SC710 | 0.69 | AN211 | 325 | bax16 | 0.11 | BD115 | 0.34 | BF157 | 0.33 | BRC6109 | 0.83 | CA3089 | 0.83 |
| 17074 | 930 | 2SB511 | 250 | 2SC711A | 0.50 | AN2140 | 240 | BC107 | 0.13 | BD116 | 0.70 | BF158 | 0.18 | BRC82 | 1.08 | CA3090aa | 325 |
| 17089 | 3.45 | ${ }^{2 S B 54}$ | 1.39 | ${ }^{2 \mathrm{SC}} \mathrm{C} 717$ | 128 | AN234 | 5.92 | BC107a | 0.11 | BD124 | 1.31 | 8F559 | 0.18 | BRC83 | 219 | CA3094 | 220 |
| 17127 | 250 | ${ }^{2 S 8546}$ | 0.56 | ${ }^{2 S C 7734}$ | 1.43 | AN236 | 3.78 | 8C1078 | 0.18 | 80124P+KIT | 0.69 | ${ }^{\text {BFI } 160}$ | 0.31 | BRC84 | 208 | CA3131EM | 295 |
| 17376 | 1.58 | 2 2S856 | 2.80 | ${ }^{2 S C 761-Y}$ | 0.95 | A 2339 | 4.68 | ${ }^{\text {BC108 }}$ | 0.08 | ${ }^{80131}$ | 0.57 | ${ }^{\text {BFI67 }}$ | 0.38 | B8X44 | 0.60 | CBE16848N- 071 | 1.56 |
| 1 14001 | 0.04 | 2SB618a | 222 | ${ }^{2 S C 783}$ | 3.98 1.85 | AN240p | 125 | ${ }^{8 \mathrm{BCL} 108 \mathrm{~B}}$ | 0.15 | ${ }^{8 D 132}$ | 0 | ${ }_{\text {BFI73 }}$ | 0.334 | BR×49 | 0.67 | CD4001 | 0.34 |
| 1N4002 | 0.06 | ${ }_{2 S 8643}$ | 0.80 | ${ }_{2 S 888}$ | 188 0.28 | AN241 | 1.71 | ${ }_{8 \text { BC109 }}$ | 0.12 | ${ }^{\text {BD }} 1335$ | ${ }_{0}^{0.35}$ | ${ }_{\text {BF178 }}$ | 0.40 | ${ }^{\text {BRY }} 39$ | 0.69 | ${ }^{\text {COU402 }}$ | 027 1.35 |
| 1/ 4003 <br> in 40004 | 0.06 | ${ }^{2}$ 2S8669 | 3.67 | $2 \mathrm{CSC867A}$ | 3.84 | AN245 | 4.49 1.80 | BC109C | 0.12 | ${ }^{\text {BDI } 136}$ | 0.26 | ${ }^{\text {BFF179 }}$ | 0.36 | BSS38 ${ }_{\text {BSTB0140G }}$ | ${ }_{5}^{0.87}$ | CD4011 | 029 |
| 1N4005 | 0.05 | 2SB681 | 335 | ${ }^{2 S C 876}$ | 0.95 | AN260 | 1.85 | ${ }^{\text {BCl13 }}$ | 0.14 | ${ }^{8 D 137}$ | ${ }^{023}$ | ${ }^{\text {BFI } 180}$ | ${ }_{0}^{0.36}$ | ${ }^{\text {BSTC0246 }}$ | 6.99 | CD4012 | 0.34 |
| 1 N4006 | 0.08 | ${ }_{\text {2SB75 }}$ | 1.98 | ${ }_{2 S C 935}^{25}$ | 4.13 | AN262 | 120 | ${ }^{8 C 126}$ | 0.33 | ${ }^{80138}$ | 0.38 | ${ }^{\text {BFI } 182}$ | 0.34 | BSTC0233 | 725 | CDU4013 CD4016 | ${ }_{0}^{0.43}$ |
| 1N4007 | 0.07 | 2 SB 774 | 0.65 | 2SC936 | 8.66 | AN272 | 825 | 8C132 | 0.14 | BD140 | 0.29 | BF183 | 0.39 | BSTCC0143 | 3.07 | CD4017 | 0.82 |
| 1 1Na148 | 0.03 | 2 28819 | 1.13 | 2SC940 | 4.68 | AN295 | 5.52 | BC135 | 0.14 | BD144 | 1.70 | BF184 | 0.43 | BSTD1043 | 2.85 | CD4020 | 123 |
| 1 N4448 | 0.05 | 2SC1034 | 6.75 | ${ }^{2 S D 1128}$ | 290 | AN301 | 2.45 | ${ }^{\text {BC }} 137$ | 0.18 | ${ }^{\text {BD } 150}$ | 125 | ${ }^{\text {BF195 }}$ | 0.39 | BSV578 | 3.49 | CD4022 | 0.39 |
| 1 1 5401 | 0.14 | ${ }^{25 C 1050}$ | 5.06 | ${ }^{2 S D 1138}$ | 0.94 | AN302 | 3.99 | ${ }^{8 C 138}$ | 0.34 | ${ }^{80157}$ | 0.67 | ${ }^{\text {BFI } 194}$ | 0.14 | ${ }^{\text {BSW68 }}$ | 0.60 | ${ }^{\text {Co4023 }}$ | 028 |
| 1N5402 | 0.15 | ${ }^{\text {2SC1096 }}$ | 1.16 | ${ }_{2 S 01273}$ | 1.56 | AN303 | 4.39 | 8C139 | 0.28 | ${ }^{\text {BDI } 160}$ | 1.50 | ${ }^{\text {BFF }} 195$ | 0.14 | BSX19 | 129 | CD4025 | 0.64 |
| 1 L 5483 | 0.16 | 2SC1104 | 3.98 | ${ }^{2 S D 1453}$ | 1.40 | AN305 | 8.95 | BC140 | 0.45 | 80163 | 0.71 | ${ }^{\text {BFI } 196}$ | 0.17 | BSX20 | 0.30 | C04028 | 0.84 |
| 1 N5404 | 0.15 | ${ }_{\text {2SCl114 }}$ | 4.54 3 3 | ${ }^{\text {2SDI52K }}$ | 264 | ${ }_{\text {A }}$ A315 | 246 | ${ }^{\text {BC141 }}$ | 0.34 | ${ }^{80165}$ | 0.62 | ${ }^{\text {BFI } 197}$ | 0.18 | BSY52 | 0.50 | ${ }^{\text {CO4 }}$ C0468 | 0.85 |
| 1N5408 | 0.35 | ${ }^{\text {2SCl1116 }}$ | 4.35 | ${ }_{\text {2S023 }}$ | 420 0.49 | AN316 | 553 | ${ }^{\text {BC142 }}$ | 0.19 | ${ }^{80} 1688$ | 0.42 | BFF198 | 0.17 | ${ }^{\text {BST79 }}$ | 0.51 | CDI047 | 1.06 |
| ${ }_{1}^{1} 1014$ | 0.04 | 2SC1124 | 1.28 | 2SD235 | 0.60 | AN320 | 5.47 | ${ }_{\text {BC1 }} 14$ | 0.09 | ${ }_{\text {BDI } 175}$ | 020 | BF200 | 0.31 | ${ }_{\text {BTI }}$ | 1.45 | C04052 | 0.75 |
| IR3403 | 5.00 | 2SC1129 | 1.65 | 2 SO 24 | 229 | AN321 | 225 | BC148A | 0.11 | B0179 | 0.45 | BF218 | 0.36 | BT119 | 1.76 | CD4066 | 020 |
| 151555 | 0.31 | 2SC1131 | 0.64 | 2 2S257 | 1.98 | A 322 | 5.85 | BC148B | 0.13 | BD181 | 0.99 | BF224 | 0.17 | BT120 | 217 | CD4069 | 029 |
| ${ }_{1}^{1 S 44}$ | 0.10 | 2SC1158 | 333 | 2 2S292 | 2.59 | AN331 | 5.11 | BC148C | 0.11 | BD182 | 0.99 | BF237 | 0.65 | BT121 | 248 | CD4070 | 0.66 |
| 1S5012A | 0.81 | ${ }_{2 S C 1162}^{2 S}$ | 0.55 | ${ }^{2 S D 313}$ | 259 | AN337 | 5.37 | ${ }^{\text {BCIL49 }}$ | 0.11 | ${ }^{8 D 183}$ | 0.99 | ${ }^{\text {BF240 }}$ | 0.17 | ${ }^{\text {BTI23 }}$ | 1.88 | CD4081 | 0.35 |
| 15921 | 0.10 | 2SC1172 | 27 | 2SD325D | 226 | A N340P | 1.17 | BC1498 | 0.13 | ${ }^{8 D 184}$ | 121 | BF241 | 0.15 | BTI51-800R | 0.89 | CD4093 | 0.72 |
| 2N1303 | 0.38 | 2SC1195 | 3.26 | 2 SD348 | 16.13 | AN355 | 5.98 | BC153 | 0.14 | BD187 | 0.53 | BF245 | 0.50 | ВП6018 | 242 | CD4511 | 1.10 |
| 2N22194 | 033 | ${ }_{2 S C 1212 A}$ | 1.97 | ${ }^{2 S D 350}$ | 520 | AN362 | 1.50 | BC154 | 0.14 | ${ }^{\text {BD } 189}$ | 0.09 | ${ }^{\text {BF245A }}$ | 0.52 | ВП8124 | 4.89 | CD4528 | 204 |
| 2 22222 | 0.38 | ${ }_{2 S C 1213}^{2 S}$ | 0.89 | ${ }_{2 S}^{2 S 0353}$ | 7.50 | AN370 | 3.95 | ${ }^{\text {BC159 }}$ | 0.36 | BD190 | 0.72 | ${ }^{852458}$ | 0.49 | BU106 | 248 | CD4556 | 1.47 |
| 2 N 2646 | 0.80 | ${ }_{2 S C 1226}$ | 1.46 | ${ }^{2} \mathrm{SD} 389$ | 2.41 | AN5010 | 5.70 | ${ }^{\text {BCII60 }}$ | 0.40 | ${ }^{80201}$ | ${ }^{0.655}$ | 8F2464 | 252 | ${ }^{\text {BU }} 108$ | 1.50 | CROZAM-8 | 1.70 |
| 2N2904 | 0.35 | ${ }^{25 \mathrm{SC}} 1293$ | 0.90 | ${ }^{2 S D} 401$ | 1.40 | AN5111 AN5120 | 292 | ${ }_{\text {BC161 }}^{\text {BC16 }}$ | 0238 | ${ }^{80202}$ | ${ }_{0}^{0.50}$ | ${ }^{85255}$ | 0 | BU109 | 268 | CV12E | 4.09 |
| 2N2905 | 0.59 | ${ }_{\text {2SC1316 }}$ | 198 | 2SO414 2S0471 | 198 | ANSI20N AN5132 | 4.50 5 | ${ }_{\text {BC169C }}^{\text {BCI }}$ | ${ }_{0} 0.16$ | ${ }^{80203}$ | 0.41 | ${ }_{\text {BF25s }}$ | 0.42 | ${ }^{\text {Buthio }}$ | 5.69 4.16 |  | 3.14 |
| ${ }^{2} 2 \times 2906$ | 0.38 | ${ }_{2 S C 1317}$ | 0.50 | ${ }_{2 S 0560}$ | 295 | AN5250 | 3.98 | BC170 | 0.16 | BD207 | 1.79 | ${ }_{\text {BF256LC }}$ | 0.82 | BU125 | 248 | Cx108 | 1248 |
| 2N2928 | 0.15 | 2SC1354 | 0.49 | 2S0588A | 236 | AN5435 | 225 | BC171 | 0.11 | BD208 | 0.34 | BF257 | 0.34 | BU126 | 1.45 | CX109 | 786 |
| ${ }^{2}$ 2N3053 | 0.35 | ${ }_{2 \mathrm{SCl} 1383}$ | 120 | ${ }^{2 S D 600}$ | 298 | AN5610 | 5.50 | ${ }^{\text {BC }} 172$ | 0.13 | BD222 | 0.50 | ${ }^{\text {BFF258 }}$ | 0.36 | BU137 | 6.53 | ${ }^{\text {Cx }} 130$ | 8.76 |
| 2N3054 | 0.99 | 2SC1394 | 245 | 2SD601R | 0.65 | AN5612 | 4.68 | BC172B | 0.27 | BD225 | 0.49 | BF259 | 0.34 | Bu205 | 1.35 | ${ }^{\text {Cx134 }}$ | 1232 |
| 2N3055 | 0.61 | 2SC1398 | 0.79 | $2 \mathrm{SD613}$ | 1.03 | AN5613 | 4.63 | BC173 | 0.17 | B0228 | 0.63 | BF262 | 028 | BU206 | 127 | ${ }_{\text {Cx }} 136$ | 11.49 |
| 2N3442 | 1.56 | ${ }^{2 S C 1413 A}$ | 3.05 | ${ }^{250521}$ | 12.85 | AN5630 | 3.95 | BC1748 | 027 | ${ }^{80229}$ | 1.05 | ${ }^{\text {BFF263 }}$ | 0.57 | BU207 | 1.60 | ${ }^{\text {cx }} 139$ | 11.83 |
| 2N3702 | 0.14 | ${ }_{2 S C 1446}$ | 125 | ${ }^{250636}$ | 0.55 | AN5701N | 1.65 | ${ }^{\mathrm{BCL} 17}$ | 0.35 | ${ }^{\text {B0232 }}$ | 0.50 | ${ }_{\text {BF271 }}$ | 0.34 | ${ }^{\text {BU208 }}$ | 120 | CX 157 <br> $\mathrm{C} \times 58$ <br> 158 | 5.52 |
| ${ }^{2}$ 2N37705 | 0.18 | ${ }^{\text {2SC1475 }}$ | 207 | ${ }^{25}$ | 0.12 | ANN250 | 295 | ${ }^{\text {BC179 }}$ | 026 | ${ }^{80234}$ | 0.47 | ${ }^{\text {BFP273 }}$ | 020 | BU20802 | 1.97 | cxis8 cıin | 5.54 |
| 2N3706 | 0.14 | 2SC1505 | 1.00 | 2 20657 | 3.50 | Ang310 | 8.74 | BC182 | 0.05 | ${ }^{8} 2238$ | 039 | BF324 | 0.35 | BU208D | 1.95 | ${ }_{\text {c }} \times 187$ | 6.84 |
| 2N3707 | 0.16 | 2SC1514 | 1.69 | 2S0661A | 0.80 | Ang320N | 428 | BC182L | 0.10 | 80239 | 0.45 | BF336 | 0.33 | BU209 | 1.50 | Cx755 | 1295 |
| 2N3711 | 0.13 | 2SC15730 | 125 | 2 2S731 | 1.05 | AN6340 | 10.14 | BC182LB | 0.01 | BD240 | 0.57 | BF337 | 0.45 | BU226 | 245 | CX885A | 6.85 |
| 2N3771 | 0.70 | 2SC1578 | 8.74 | 2 SD773 | 0.60 | An6341 | 298 | ${ }^{\text {BC1 } 1832}$ | 0.11 | ${ }^{\text {BD2 }}$ 21 | 3.39 | BF338 | 0.33 | BU326 | 200 | DEC1 | 220 |
| 2N3772 | 1.71 | 2SC1583 | 0.50 | 2 SD811 | 3.30 | AN6342 | 27 | BC183LB | 0.2 | 80242 | 0.39 | ${ }^{\text {BFF355 }}$ | 0.99 | BU326A | 220 | DEC2 | 220 |
| ${ }^{2} \times 3773$ | 1.65 | ${ }_{2 \text { SC1617 }}$ | 3.89 | ${ }^{2 S 0823}$ | 1.98 | AN6363 | 16.00 | BC184 | 0.13 | BD243A | 0.35 | ${ }^{\text {BF362 }}$ | 0.62 | BU326S | 220 | DS3486N | 4.33 |
| 2N3819 | 0.54 | 2SC675 | 1.41 | 2SD837 | 1.56 | AN6371 | 924 | BCi84L | 0.14 | BD243C | 0.29 | BF363 | 0.50 | BU406 | 1.49 | DS3487N | 4.95 |
| 2N3823 | 1.17 | 2SC1678 | 1.98 | ${ }_{2} 2$ SD841 | 2.50 | AN6387 | 10.65 | BC184LB | $0 \times 1$ | 80244 | - 4 | ${ }^{\text {BFF371 }}$ | 0.50 | BU4060 | 1.79 | E1222 | 0.40 |
| 2N3904 | 0.62 | ${ }^{2 S C 1741}$ | 127 | ${ }^{2508565}$ | 1.00 | AN6531 | 1.95 | ${ }^{\text {BC }} 1886$ | 027 | ${ }^{80244 C}$ | ${ }^{1.79}$ | ${ }_{\text {BF391 }}^{\text {BF917 }}$ | 025 | BU407 | 0.82 | ${ }^{\text {E5522 }}$ | 0.28 |
| ${ }^{2} \mathrm{~N} 3908$ | 0.62 | 2SC1810 | 1.70 | 2S08570 | 1.84 | AN6551 | 1.35 | ${ }^{\text {BC187 }}$ | 0.28 |  | 1.99 | ${ }^{\text {BF4 } 47}$ | 0.84 | BU4070 | 0.99 | E5386 | 025 |
| 2N4101 | 1.3 | 2SC1815 | 0.45 | 2 2SD882 | 1.15 | AN6552 | 0.68 | ${ }^{\text {BC204 }}$ | 0.16 | 8D246C | 0.7 | BF418 | 1.87 | Bual2 | 529 | E9003 | 0.46 |
| ${ }^{2} \mathrm{~N} 424400$ | 3.30 0.99 | ${ }^{\text {2SCC1826 }}$ | 0.67 228 | 2SD894 2S098 | 1.75 | AN6610 AN6677 | 240 | BC207 BC212 | ${ }_{0}^{0.14}$ | BD253 BD2784 | ${ }^{1.05}$ | ${ }_{\text {BFent }}^{\text {BF422 }}$ | 029 | BU426A BU500 | 1.13 | ${ }_{\text {E }}^{\text {E9005 }}$ | 0.50 588 |
| 2N5293 | 0.50 | 2SC1875 | 4.50 | ${ }^{2 S K} 105 \mathrm{H}$ | 215 | AN7111 | 125 | BC2128 | 0.26 | BD317 | 260 | BF450 | 0.35 | BU508a | 125 | GC374 | ${ }_{1.65}$ |
| 2N5294 | 0.50 | 2SC1881K | 2.38 | 2SK152 | 3.59 | AN7114E | 8.54 | BC213L | 0.10 | BD318 | 200 | BF451 | 029 | BU536 | 1.65 | G0243 | 4.34 |
| 2N5296 | 0.49 | 2SC1893 | 3.02 | 2SK34 | 0.76 | AN7115 | 3.38 | BC213LB | 0.15 | 80375 | $0_{42}$ | BF457 | 0.41 | BU608 | 1.80 | G7758 | 0.84 |
| ${ }^{2 N 5297}$ | 0.50 | ${ }^{2 S C 1906}$ | 0.98 | 2SK41 | 1.07 | AN7120 | 4.65 | ${ }_{\text {BC2 }}{ }^{\text {B } 214}$ | 0.10 | 80330 | 0.76 | ${ }^{\text {BF458 }}$ | 0.35 | BU705 | 295 | GH3F | 1.82 |
| 2N5298 | 0.61 | $2{ }^{\text {SCl } 1921}$ | 1.37 | 2SK79 | 298 | AN7145 | 280 | BC214LB | 0.26 | 80410 | 0.52 | BF459 | 0.52 | BU806 | 1.79 | HA11215 | 1.75 |
| 2N5771 | 1.18 | 2SC1923 | 0.30 | 40408 | 0.50 | AN7146 | 4.35 | BC225 | 0.40 | ${ }^{80433}$ | 047 | BF460 | 1.45 | BU807 | 0.80 | HA11219 | 253 |
| ${ }^{2} \mathrm{~N} 6109$ | 158 | ${ }^{2 S C 1929}$ | 225 | 40594 |  | AN7151 | 226 | ${ }^{\text {BC237 }}$ | 0.10 | ${ }^{80434}$ | 049 | BF649 | 0.25 | BU826a | 1.95 | HA11225 | 1.50 |
| 2N6130 | 0.80 | 2SC1942 | 1.65 | 40636 | 1.43 | AN7156 | 285 | ${ }^{\mathrm{BC} 2378 \mathrm{BJ}}$ | 0.12 | ${ }^{80435}$ | 049 | BF470 | 0.55 | BUW84 | 1.39 | HA11226 | 10.44 |
| 2N6133 | 125 | 2SC1945 | 7.99 | 4EX581 | 0.80 | AN7158 | 238 | ${ }_{\text {BC2 }}{ }_{\text {BC } 238}$ | ${ }_{0}^{0.13}$ | ${ }^{\text {BDa36 }}$ | 0.60 | ${ }_{\text {BFP47 }}^{\text {BFA }}$ | 033 | BUX84 | 1.00 | HA11229 | 1.96 |
| ${ }_{2}$ N66292 | 1.95 | ${ }_{2 S C 1957}$ | 1.09 | 7805-T022 | 0.30 0.63 | ${ }_{\text {an }}^{\text {ant223 }}$ | 4.25 | ВС238B | 0.08 | B0438 | 040 | BF479 | 0.35 | BuY69a | 204 | HA11124 | 5.25 |
| 2N696 | 0.43 | 2SC1953 | 1.93 | 7806 | 0.73 | AU107 | 3.50 | BC239 | 0.12 | BD441 | 1.42 | BF480 | 1.38 | BY126 | 0.13 | HA11244 | 4.02 |
| ${ }^{2} \mathrm{~N} 698$ | 0.43 | 2SC1962 | 1.93 | 7808 | 0.85 | AUl10 | 225 | BC2398 | 025 | BD442 | 1.41 | ${ }^{\text {BFP491 }}$ | 1.99 | ${ }^{\text {BY127 }}$ | 0.08 | HA11251 | 4.47 |
| 2 2SA1006 | 1.50 | 2SC1969 | 204 | 7812-T022 | 0.35 | AU113 | 525 | BC251A | 0.31 | BD509 | 1.55 | BF995 | 0.64 | BY133 | 0.12 | HA1125 | 429 |
| ${ }^{\text {2SA }}$ 2SA1011 | 1.65 | 2SC1983 | 1.51 | 7815 | 0.64 | AY105K | 208 | ${ }^{\text {BC294 }}$ | 0.50 | ${ }^{\text {BD5 }} 10$ | 0002 | ${ }^{\text {BF506 }}$ | 0.43 | ${ }^{\text {BY1 }} 164$ | 0.45 | ${ }_{\text {HA1 }}$ | 4.87 |
| 2SA1012 | 125 | 2SC2009 | 0.34 | 7824 | 0.64 | A ${ }_{\text {A }}$ | 881 | ${ }^{\text {BC }} 301$ | 0.45 | ${ }^{\text {BDO529 }}$ | 0.38 | ${ }^{8}$ | 0.24 | ${ }^{\text {BY179 }}$ | 1.08 | HA11414 | 5.15 |
| 2SA1020\% | 0.89 | 2SC2029 | 233 | 7905 | 0.80 | 8250 | 225 | BC302 | 0.53 | 80530 | 1.18 | BF532 | 0.45 | BY182 | 0.95 | HA1144 | 7.87 |
| 2SA1027R | 0.45 | 2SC2028 | 211 | 9368 | 10.70 | B40 | 1.55 | BC303 | 1.04 | BD533 | 0.15 | BF596 | 0.18 | BY184 | 0.37 | HA1156 | 1.16 |
| ${ }^{254} 473$ | 0.75 | 2SC2063 | 0.99 | AA133 | 0.12 | BA130 | 0.14 | BC307 | 0.18 | BD534 | 0.53 | BF597 | 027 | BY187 | 0.7 | HA1160 | 4.78 |
| ${ }^{\text {2SAF }}$ 2S5S ${ }^{\text {2 }}$ | 4.95 | 2SC2078 | 3.11 | ${ }_{\text {ACl }}$ | 0.12 | ${ }_{\text {BA } 13710}^{\text {BA }}$ | 1.98 | ${ }^{\text {BC3307A }}$ | 008 | ${ }^{80535}$ | 0.79 | ${ }^{86694}$ | 0.22 | ${ }^{\text {BY189 }}$ | 1.79 | HA11166 | 1.90 |
| ${ }_{\text {2SCl173Y }}$ | 125 | ${ }_{\text {2SC2085-a }}^{\text {2S }}$ | 225 | ${ }_{\text {ACl27 }}{ }^{\text {ACl23K }}$ | 0.43 0.27 | BA 1320 BA 132 | 1.38 <br> 3.95 | ${ }_{\text {BC }}$ | 0.18 | ${ }_{80537}^{8053}$ | 0.0 .61 | ${ }^{87757}$ | 0.0 .59 | ( ${ }_{\text {BYY98 }}^{\text {BY2012 }}$ | 1.62 | HA1166X | ${ }_{5}^{6.43}$ |
| 2SC1509 | 1.35 | 2SC2091 | 1.30 | AC128 | 0.34 | bal 1330 | 275 | BC309 | 0.17 | B0538 | 0.80 | BF761 | 1.05 | BY20320 | 0.59 | HA11706 | 3.61 |
| 2SD1391RL | 3.95 | 2SC2141 | 244 | ${ }^{\text {AC138 }}$ | 0.24 | BA 145 | 0.19 | BC317A | 0.13 | B0544B | 0.83 | BF762 | 0.50 | BY207 | 0.22 | HA17705 | 8.00 |
| ${ }^{25 A 1095}$ | 3.00 | 2SC2166 | 1.98 | AC141 | 029 | BA 148 | 0.25 | BC327 | 0.15 | BD598 | 1互 | BF869 | 0.47 | BY208 | 0.46 | HA11703 | 4.22 |
| ${ }_{2}$ SA1103 | 6.55 | ${ }^{2 S C} 2816$ | ${ }^{0} .69$ | ${ }^{\text {ACCl42K }}$ | 0.35 | ${ }^{\text {BA } 154}$ | 0.40 | ${ }^{\text {BC328 }}$ | 0.10 | ${ }^{80677}$ | 0.0 | BF870 | 0.30 | ${ }^{\text {BY } 210.400 ~}$ | 0.19 | Hallivol | 4 |
| 2SA329 | 0.40 | ${ }_{2 S C 2233}$ | 180 | AC151 | 028 | ${ }^{81} 155$ | 0.12 | ${ }^{\text {BC337 }}$ | 0.09 | ${ }^{80679}$ | 0.5 | ${ }^{\text {BF959 }}$ | 0.42 | BY210-600 | 027 | HA171710 | 9.50 |
| 2SA990 | 225 | 2SC2278 | 1.69 | ${ }_{\text {ACl79 }}$ | 0.28 | BA159 | 0.08 | ${ }_{\text {BC368 }}$ | 0.24 | ${ }^{806581}$ | 1.48 | ${ }^{\text {BF970 }}$ | 0.49 0.50 | ${ }^{\text {BY }}$ 8210-800 | 0.30 <br> 1.64 | HA117 HAlli | 9.75 20.16 |
| ${ }^{2 S A 493}$ | 225 | 2SC2214 | 217 | AC183 | 0.72 | BA182 | 024 | BC440 | 0. © | BD696 | 247 | BFF33 | 0.44 | BY223 | 123 | HA11715 | 325 |
| 2 2SA562 | 0.57 | ${ }_{2 S}^{2 S C 2335}+\mathrm{KIT}$ | 13.4 | ${ }^{\text {ACP187 }}$ | 0.39 | BA232 | 1.66 | BC441 | 0.44 | ${ }^{80699}$ | 3.49 | BFR61 | 0.92 | BY224600 | 1.88 | HA11714 | 9.75 |
| ${ }^{2 S A 564}$ | 0.75 | ${ }^{2 S C 2551}$ | 123 | ${ }^{\text {ACTi87K }}$ | 0.43 | BA302 | 124 | BC454 | 0.36 | BD700 | 370 | BFAB62 | 0.50 | BY225-100 | 1.13 | HA17716 | 13.10 |
| ${ }_{2}^{2 S A 614}$ | 4.88 | 2SC2565 | 3.92 | AC138 | 0.37 | ${ }^{\text {BA3 }} 1$ | 1.35 | ${ }^{\text {BCa }}$ 80 | 0.42 | ${ }^{80777}$ | 0.90 | BFR79 | 029 | ${ }^{\text {BY226 }}$ | 023 | HA11725 | ${ }^{1828}$ |
| ${ }_{\text {2SA6393 }}$ | 1.15 | ${ }^{\text {2SC2570 }}$ | ${ }_{2}^{288}$ | ${ }^{\text {AC }}$ AC1888K | ${ }_{0}^{0.44}$ | ${ }_{\text {BA }}^{\text {BA312 }}$ | 1.45 0.76 | ${ }^{\text {BC461 }}$ | 0.35 | ${ }^{\text {BD709 }}$ | 1.05 | ${ }^{\text {BFRP1 }}$ | 1.6 | ${ }_{\text {BY27 }}^{\text {BY27 }}$ | 0.20 | HA11725MP HA11755P | 16.00 |
| 254659 | 0.49 | 2SC2578 | 6.73 | AC193K | 0.65 | BA317 | 0.08 | BC463 | 0.64 | BD809 | 0.94 | BFR89 | 1.63 | BY229-1000 | 1.12 | HA11781 | 19.90 |
| ${ }^{25 A 673}$ | 1.50 | ${ }^{2 S C 2671}$ | 1.99 | ACIgak | 0.65 | BA318 | 0.02 | ${ }^{\text {BC477 }}$ | 0.37 | B0810 | 0.69 | bfrgoa | 0.70 | BY229-600 | 0.92 | HA1180 | 5.15 |
| 2SA684 | 1.61 | ${ }^{2 S C 28286}$ | 207 | AD140 | 1.06 | BA328 | 1.05 | ${ }^{\text {BC478 }}$ | 0.21 | BD879 | 0.76 | BFI42 | 0.43 | BY255 | 0.66 | HA1196 | 1.43 |
| 2SA697 2SA 699 | 1.05 | ${ }_{2 S} 252888$ A | 1.85 | AD143 | 1.93 | BA333 | 1.37 | BC479 | 0.41 | ${ }^{\text {B0880 }}$ | 0.78 | ${ }^{\text {BrF43 }}$ | 0.43 | BY295-600 | 1.03 | HA13001 | 1.73 |
| 2SA715 | 0.95 | ${ }_{2 S}{ }^{\text {Sc372 }}$ | 1.40 | ${ }^{\text {ADP }} 161$ | 1.60 0.30 | ${ }_{\text {BA }}$ | 286 | ${ }^{\text {BC5 } 546}$ | 0.08 | ${ }^{\text {BDP899 }}$ | 245 | ${ }_{\text {BFW } 10}^{\text {BF }}$ | 0 | ${ }^{\text {B7298 }}$ | 0.36 0.45 | HA1306 HA1338 | 225 |
| 254747 | 10.74 | ${ }^{25 C 373}$ | 1.16 | A0162 | 0.30 | BA511 | 1.95 | BC547 | 0.10 | BD901 | 0.79 | BF29 | 0.34 | BY407 | 0.90 | HA1339 | 3.40 |
| ${ }_{2 S A 848}$ | 1.36 | 2SC333 | 1.33 | AD262 | 125 | BA514 | 220 | BC548 | 0.10 | B0902 | 0.84 | BFX84 | 0.37 | BY409 | 1.49 | HA13402 | 787 |
| ${ }_{2 S A 8 B 5}$ | ${ }_{2}^{0.05}$ |  | 0.50 | ${ }_{\text {AF115 }}^{\text {AF } 14}$ | 2.78 | ${ }^{\text {BA521 }}$ | 88.94 | BC549 BC550 | 0.10 0.10 | ${ }_{\text {BDWB4C }}$ | 1.45 1.56 | - | 0.31 |  | 135 0.65 | HA 13342 HA 13365 | 265 4.02 |
| 2 24a33 | 0.99 | 2SC403C | 0.60 | AF118 | 120 | BA566 | 7.98 | ${ }^{\text {BC556 }}$ | 0.10 | BDX32 | 1.75 | BFX87 | 0.55 | BYW19:1000 | 0.69 | HA1366WR | 1.50 |
| 254884 | 0.65 | ${ }^{2 S C 41}$ | 2.19 | AF127 | 0.79 | BA527 | 2.98 | ${ }^{\text {BC55 }}$ 7 | 0.10 | 8DX53a | 125 | ${ }^{\text {BFX } 888}$ | 0.34 | ${ }^{\text {BrW5 }}$ | 0.16 | HA1367 | 275 |
| ${ }_{2 S A 872}$ | 0.80 | 2SC458 | 0.15 | AF139 | 0.40 | BA532 | 1.50 | BC558 | 0.10 | BDX538 | 1.85 | BF889 | 0.44 | B B 10 | 029 | HA1368R | 245 |
| ${ }_{\text {2SAA937R }}$ | 215 | ${ }_{2 S}^{2 S C 495}$ | ${ }_{2}^{0.98}$ | ${ }_{\text {AFF }}^{\text {AF7 }} 178$ | ${ }_{0}^{1.45}$ | ${ }_{\text {BAF } 209}^{\text {B }}$ | 205 | ${ }_{\text {BC5599 }}$ | 0.10 | ${ }^{\text {BDX }}$ B698 | 216 | ${ }_{\text {BFY }}$ | 0.38 | 8YX55-600 | 023 | ${ }_{\text {HA1368 }}$ | 207 |
| IF YOU |  | 2 C515 ${ }^{\text {a }}$ |  | Aflig |  | BA6209 | 4.55 | ВС5998 | 0.11 | B0X624 | 215 | BFYS | 025 | BY771-600 | 0.85 | HA1370 | 3.30 |



## The Art of Servicing

B.A. Berry

As an old hand at this radio and television servicing business it saddens me that fault finding is becoming a lost art. The growing use of chips is doubtless responsible for much of this lack of finesse, and as ever more advanced techniques are coming into use the situation is getting worse. There's still room for the art of fault finding however, even if it's only in fault location to panel level. Nowadays, on being handed a piece of faulty equipment to repair, too many youngsters charge straight in without thinking. Even when they do think they invariably assume that the fault is the most complicated one they can imagine. The next time you're handed a piece of equipment for repair - stop right there and think! Nearly all faults can be isolated to a particular area without bringing even a test meter into use.

## Customer Interrogation

Before you let the customer out of the shop it's imperative that you give him a real third degree on just what's gone wrong and how it happened. Write this down, because the moment he's gone you'll forget the .most important point. It's also extremely important that you ask whether he attempted to do anything about the fault himself. Most people will cheerfully leave say a camera repair job to a specialist but will quite happily take a screwdriver to their video recorder or TV set, then expect you to be able to diagnose and repair a fault when the equipment has been misaligned. It's happened to me all too frequently.

So take careful notes before the customer leaves. The sorts of things you should ask are: At what time and where did the fault occur? Was the customer in the room at the time? For how long had the equipment been working satisfactorily? What are the symptoms, and was there any smoke or peculiar noises? What action did the customer take when the fault occurred, and did the equipment get hot? If so, ask him to point out the exact spot on the external case. Had the customer been doing anything to the equipment at the time? You'd be surprised for example how many people attempt to join up a speaker extension lead while the equipment is working!

The final thing before the customer leaves is to ask him tactfully whether or not he's attempted any repair action himself. Point out that this question is merely to save him money in the long run. If any trimmers or preset controls have been turned, find out now. Such information could cut down the servicing time considerably. Much as you would like to read the riot act at this stage, don't. To do so would only result in the customer denying that he'd even thought of touching anything.

## Preliminary Assessment

Having got the equipment on the bench you may feel that you are now justified in removing the outside case. Not so! If you are unfamiliar with it, get the service manual out - if you have one - and check on the system and circuit configuration. In a great many cases, especially with the more sophisticated types of equipment that are being increasingly brought in for attention, the cause of the trouble can simply be a matter of misadjustment, e.g. a

TV/aux switch in the wrong position. It's so easy for even an experienced engineer to miss the obvious when questioning a customer in a perhaps crowded shop. So unless the reported fault obviously requires a look inside, don't unbox the equipment. Instead, apply power and commence your own investigation by checking all external control settings and indicator lamps. Don't neglect input fuses, even when the customer has told you that he changed the fuse. Some 3A and 13A mains fuses are notoriously unreliable. I always make a habit of putting the meter across a new one just in case.

As you go through the various controls, note exactly what they do or don't do. Take your time over this, because this is the stage at which you will be forming your own opinions as to the likely cause of the fault. I always remember the advice given to me by an old engineer who taught me the trade: eyes first, cars next, fingers last of all! It's stood me in good stead over the years.

## Initial Checks Inside

By now you will have formed a preliminary idea of what the problem is and where the cause lies. So power off and unbox. At this stage the most useful tool may well be a large magnifying glass. A thorough and concentrated look at the board and the components on it may well reward you with an easy repair. Cracks, solder bridges, overheated components and dry-joints are easy to see under a lens.

With equipment that's been operating satisfactorily for at least six months component failure is only rarely the cause of a fault - unless the component has been subjected to outside influences! Modern components are very reliable. Remember the bathtub curve which clearly shows that most component failures in solid-state equipment occur in the first few months. Failures then fall to a very low level for the normal life span of the equipment, rising again as the equipment reaches the end of its expected life span. During the long period between the initial burn in and old age most faults are due to the causes previously listed, with dry-joints leading the list of possible culprits. In the main they can be found quite easily with a lens.

## Test Equipment Next

Component failures do of course occur from time to time in otherwise healthy equipment. It's then that you need the meter or scope. I can't emphasise too much the wisdom of measuring supply voltages and currents. They can give a very good clue to the cause of a fault - especially if you've been clever enough to measure these voltages and currents in a similar piece of equipment that's working. Yes, I know that the readings are given on circuit diagrams - but not always, and not always the particular ones you want. It pays to make your own measurements and keep a note of them. A rise in supply current will lead you to look for a short-circuit, while a reduction should lead you to a burnt out or open-circuit component.

After checking the supply voltages make voltage checks around the transistors in the suspect area. The fastest way to check a transistor in an amplifier circuit is to measure its base-emitter bias - with a normal silicon transistor the reading should be around 0.7 V . A quick front-to-back
resistance check on any diodes in the fault area comes next. If a diode reads o.k. but you're still suspicious, change it. I've met some really nasty diodes in my time - ones that check out fine but prove, on replacement, to have been the cause of the equipment failing to work correctly.
If everything else seems to be in order it's time to suspect the i.c.s on the board. Undoubtedly the fastest method of checking is by substitution, particularly with some of the special devices that are around today. If you haven't got a replacement to hand you might find that there's a second chip of the same type on the board. This can be swapped over with the suspect to see whether a different fault appears. If so, you've found your culprit.

I've found that the little RS logic testers that can be clipped over a chip are very handy, though somewhat expensive: LEDs give an instantaneous indication of the logic state at each pin, making the job much easier. Don't neglect the old-fashioned signal tracer with audio equipment. A quick probe around with one of these can locate the source of a fault in a matter of minutes.

## DC Amplifiers

About the worst type of fault I've met in audio equipment, at least of the older type, is where there are several d.c. coupled amplifier/driver/output transistors. When one of the output transistors goes short-circuit normally one or more of the other transistors fails with it. If you try to work out what's wrong by taking voltage readings in a logical manner you can find yourself running round and round in circles. The best approach is to start with the first transistor involved. Remove and check it - the ordinary Avo tests will do nicely. If necessary, replace it. But don't switch on again until you are sure that all the transistors and diodes in the circuit are o.k. The amount of distress this procedure will save makes the time taken well worthwhile.

## Don't Twiddle Coils

Perhaps the best advice I can give the up and coming engineer on TV repairs is not to assume that any coils are misaligned. The occasions when this is the case are rare indeed - unless the customer has been at it! I can well recall the grief I caused myself in my earlier years by assuming that a twitch here and there would provide a cure - only to discover that it didn't, and eventually that the cause of the trouble was a dry-joint, leaving me with an unnecessary realignment job - it was this, by the way, that led my boss to give the little lecture referred to earlier! I'm well aware that there are those of you out there who consider yourselves to be perfectly able to align a TV set by eye and ear. The next time you try it, have a look at the $3 \cdot 5 \mathrm{MHz}$ bars in a test pattern - they won't be a pretty sight!

## Electromechanical Equipment

With any equipment that employs both mechanical and electronic techniques the cause of trouble is much more likely to lie in the mechanical side. Where an electronic component is subject to wear, this will probably be the cause of its failure - video and audio heads provide clear examples of this. And we all know the problems that the tape path can cause with VCRs. It's worth emphasising again that with this sort of trouble it's your eyes that will be of most use to you: study the problem until you are quite sure of what is causing the fault, and only then start stripping the equipment down.

## next month in



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Faclt-finding notes covering mainly the HS3C3, HS320, HS304, HS33J, HS306 and HS307. Written up by Derek Snelling on the basis of his experience witn these machines on rental and sales contracts.

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## - ALL ABOUT BAR CODES

Bar code scanning $r$ as for some time been used for pricing in shops and has been taken up as a simple means of programming domestic electronic equipment. Harold Peters explains how the coding system works.

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## Amstrad 4600

The capstan servo lost lock when the machine had warmed up. A slight problem was that the main servo chip IC303 (BA718) responded to freezer but the fault was still there after it had been replaced - not to mention the hassle of getting one . . The culprit turned out to be the operational amplifier chip IC302 which drives IC303 and is next to it. This device had obviously caught some of the freezer. S.B.

## Grundig VS380

The problem with this machine was no clock display. The clock and calendar counter, with serial clock and data outputs to the main microcomputer, is IC2080. All voltages seemed to be o.k. but the 32 kHz quartz crystal was very quiet. So the crystal and IC2080 were replaced - to no effect. After much deliberation it was realised that the 5 V reading at pin 15 (VSS1) was incorrect due to the serial ni-cad cell being short-circuit. The correct voltage is 3.8 V . Replacing the cell restored the clock operation. All other functions had to be reset using the service/pause function to preload the RAM.
S.B.

## Toshiba V5470

The playback picture looked as though the TV set's field oscillator was running fast - locked at twice field frequency with two partially superimposed pictures. The oscilloscope showed that part of the video signal was missing, but only on alternate fields. As the f.m. playback signal was full and correct it wasn't head or mechanical alignment trouble.
Attention was turned to the muting circuits on the servo/logic panel where C608 was found, by substitution, to be low in value. It's in the muting hold-off circuit and prevents muting as long as the control track pulses are present. In this case the capacitor's changed value meant that the time-constant wasn't long enough.
S.B.

## Amstrad 5200

There was no operation with this machine. Q651 in the 18 V power supply was open-circuit and the supply had a short to chassis across it. This was traced to the BA718 chip IC307 which was short-circuit. Replacing the chip and the transistor restored full operation.
S.B.

## Panasonic NV370

There was no playback colour - the a.p.c. loop was not locking as the reference frequency was way off. The PCB module component was replaced.
S.B.

## Panasonic NV100

Quite an old machine, this one. It had two faults, no playback audio and no playback colour. Lack of audio was due to a leaky capacitor, C4016-I knew it was leaky because the voltage at pin 1 of IC4001 was low at about 2 V instead of 4.3 V . The absence of colour was more of a problem - IC8001, a hybrid device, had been changed. It took quite a long time to measure all the frequencies and set up the a.f.c. and a.p.c. circuits. The result of all this effort was no colour for the first few seconds, the exact time depending on how long the machine had been switched off. After an overnight rest it was two-three minutes whereas after a half-hour off period the colour stayed away for only
thirty seconds. It did make fault-finding tiresomely long, but the customer wanted it done. Anyway, the a.f.c. loop was found to be locked during the monochrome period so this was cleared of suspicion, leaving the a.p.c. loop. In fact the ident output at pin 41 of the relevant chip was highly active, confirming the diagnosis. None of the frequencies were off and I was getting down to change the i.c. when I spotted C8009. Replacing this cured the problem.
S.B

## Grundig VS180

If you find that one of the reel motors is running continuously check C301 or C305 in the motor drive circuit for being leaky.
S.B.

## National Panasonic M5 Camcorder

If the complaint with one of these machines is noisy audio on playback of its own recordings, before spending a lot of time searching for some obscure fault in the microphone/ audio record section first try checking the d.c. lead to the drum motor. I've found that the ribbon cable supplying the d.c. presses against the lower drum motor, causing audio pickup of a clicking noise which is transferred to the microphone.
K.K.

## Fisher FVHP530

The fault report was no channel operation - you couldn't get any test signal on the monitor either. All 1.t. outputs from the power supply were checked. A $3 \cdot 15 \mathrm{~A}$ fuse was found to be open-circuit as a result of Q907 (B698) being short-circuit. After checking transistor data a BD234 was fitted as a replacement. This along with a new fuse solved the problem.
R.S.N.

## JVC HRD120/Ferguson 3V35

A good rule of thumb with microcomputer based mechacon and syscon circuits is that the main microcomputer chips themselves seldom fail. Before replacing them check the relevant d.c. lines, the clock pulses, etc., and remember that the various buffer chips are more prone to failure than the microcomputer chips.
We've recenlly had two HRD120s that proved to be exceptions to the rule. In the first the machine operated normally for about ten minutes before jumping into the timer mode, after which the machine became totally nonfunctional. Liberal squirts of freezer didn't have any effect. We found that in the fault condition the input to the CPU chip from the timer switch at pin 35 wasn't activated: the conditions at the rest of the CPU's pins appeared to be correct, with the trains of pulses on data line pins $27,28,29$ and 30 showing some activity on the scope. After removing half a tube of Japanese Evostick from the chip's 52 pins and fitting a replacement the machine worked normally.
The second machine couldn't be switched on by the operate switch. Again the conditions at the CPU's pins all appeared to be correct, with the level at pin 36 changing when the operate switch was selected, but this time there was a distinct lack of activity on the four data lines that connect to the input/output expander IC202. A new CPU chip was again the answer.

As an aside, we've found that the capstan motors in quite a number of these machines have become noisy. This normally has no effect on the quality of the picture and sound, and when the owner is confronted with the price of a replacement motor he's usually prepared to live with the noise. If a particularly bad motor is run in the play mode for a few hours however check that it still has sufficient torque to perform tape unloading properly -a loop of tape can be left outside the cassette and this will be damaged when the cassette is ejected.
C.H.

## Ferguson 3V20 Camera

The electronic viewfinder took a long time to display an image and when the c.r.t did light up there was lack of width with foldover on the right-hand side of the screen. The cause of the fault was traced to the line output transistor's $10 \mu \mathrm{~F} / 16 \mathrm{~V}$ base drive coupling capacitor. Inspection revealed that it had a corroded leg. A similar type and value electrolytic is used in the viewfinders field timebase circuit: this was also corroded, though there was no field fault. Both electrolytics were replaced.
A.D.

## Ferguson 3V35 and variants

Intermittent cassette loading with these machines is usually due to the insert detect switches. These switches can also cause the following symptoms: excessive force has to be used to insert a cassette, or the cassette is taken in half way then ejected. To prove whether the switches are faulty, check at pin 6 of CN27. This pin should go low while the cassette is being taken in. If the pin goes high the switches are faulty. Needless to say both should be replaced.
To operate these machines with the cassette housing unplugged and removed from the machine, switch off at the front, connect pin 5 of CN27 to chassis, then switch on. The microcomputer now thinks that the cassette housing is lowered: all functions with the exception of record can be selected and the shorting link removed. For record, connect pin 7 to chassis and select play and record as usual.
The capstan motor can be a source of rumbles which can usually be cleared with a drop of oil. To prove that the capstan motor is the source of the noise, play a tape and then select pause. If the noise stops the capstan motor is at fault. It should have no play on its shaft at all. If, by holding the motor, the pulley can be moved back and forth there's wear in the bearing. Another check worth making is to look for metal filings on the ASM board. Their presence indicates considerable wear in the bearing. Replace the motor if these checks indicate that the bearing is worn.

The video heads in these machines seem to be prone to early failure. The refurbished heads available work well. In setting up the $Q$ you might find that the trimmers are noisy.
A.D.

## Mitsubishi HS318

Not a very old machine this one - still under guarantee in fact. It would accept a cassette quietly, but when it was asked to play or record its efforts to load the tape were accompanied by a mechanical bang-bang-click effect as the loading arms jumped about violently. Moving from stop to fast forward or rewind was also a noisy business.

Inspection showed that plastic gear 1 (part no. 641D71001) had several teeth broken off it. Phasing up the mechanics after fitting a replacement can be difficult! The key to success is to align the scribe marks on the two sliders at the front (underside) of the deck, and closely follow the manual's instructions for refitting the mode
switch. Both must be done with the mechanics in the stop (not eject) position.
E.T.

## Panasonic NV8600

These oldies were built well! Some look set to go clunking and twanging into the nineties. One we had in for repair wouldn"t play or record because the pinch roller solenoid wouldn't pull in. The solenoid would hold in when operated by hand and we found that the pull-in transistor Q622 was open-circuit. A BD139 transistor turned out to be a successful replacement, but we also checked the damping diode (D624) as a precaution.
E.T.

## JVC HRD150

The owner of this machine must have had super-sensitive hearing - or a shelf or trolley that acted as a sounding board! He complained of a barely perceptible clonking noise in the record and playback modes. In a very quiet part of the workshop we could hear it: the noise was coming from the area of the supply spool turntable. We found that the supply reel clutch pinion was very slightly eccentric. This was proved by watching and listening while we spun the supply reel by hand with the back-tension band slackened. A replacement clutch assembly eliminated the trouble.
E.T.

## Panasonic NVM5B Camcorder

This was the first camcorder we've serviced without removing its case - the trouble was in the viewfinder, which clips on and plugs in. Its little screen was brilliantly lit up, with not a vestige of a picture. The diddy little monochrome c.r.t. is grid modulated by a single transistor video amplifier which is supplied by a negative line derived from pin 5 of the tiny line output transformer. The negative supply was missing because the rectifier diode was opencircuit. It's encapsulated within the transformer, so the entire unit had to be replaced - fortunately under guarantee. The replacement came in a tiny parcel . . . E.T.

## Panasonic WVP200E Camera

We and our customer almost came to blows over this old camera! He said it sometimes lost sound. We ran it for days on end with complete sound continuity. He finally convinced us by bringing in a tape recorded by the camera. There were long periods of silence, often triggered by movement of the camera. When the sound went the action of the audio a.g.c. circuit brought up the background noise, so we decided that the trouble was not far from the microphone.

And so it was! The signal from the camera-mounted microphone passes through a switch on the jack socket for the left (mono) external microphone. The socket's connecting pins were dry-jointed to their mini-PCB. E.T.

## Mitsubishi HSC20/JVC GRC7 Camcorder

The symptom with this camcorder was no threading. It would try to do so, then shut down with various function lights flashing. While dismantling it we noticed that one of the four cassette lid screws was missing. When we'd got it completely to bits we discovered that the missing screw was lodged in the loading mechanism - this was the cause of the trouble. We were much miffed to note that a good shake and rattle session would have dislodged the screw without the need to take the machine apart .

# Dual-channel TV Sound Systems 

Geoff Lewis, B.A., M.Sc.

Last month we looked at some of the basic techniques used in digital sound systems. In this concluding instalment we'll consider some of the systems in use or proposed for use in dual-channel TV sound applications.

## Dolby ADM System

The Dolby adaptive delta modulation (ADM) system is a variant of delta modulation using one bit per sample to indicate whether the analogue audio signal has increased or decreased in amplitude. It's a most effective bit rate reduction technique, allowing the use of a considerably higher sampling frequency. This in turn leads to a simpler decoder filter arrangement, without the risk of aliasing. Unlike pulse-code modulation, a single bit error has the same effect wherever it occurs. When an error bit is detected in a delta modulation system, introducing an opposite polarity bit will reduce the audible effect to almost zero. The only major disadvantage is that an overload can arise when the signal amplitude changes by an amount greater than the quantizing step size. The ADM system devised by Dolby Laboratories Inc. and adopted for use with the Australian DBS service, which uses the B-MAC transmission standard, employs both a variable step size and variable pre-emphasis to produce very high quality audio.
A pre-emphasis circuit at the encoder continuously monitors the signal frequency spectrum to determine the optimum pre-emphasis characteristic. After pre-emphasis the signal passes through a step sizing circuit which continuously evaluates the signal slope to select the best value. The pre-emphasis and step-size information is then coded as two low bit rate control signals. The audio signal is delayed by 10 msec relative to the control signals: this ensures that the control signals reach the decoder in time to enable it to decode the received audio signal in a complementary manner.

For transmission the digital signal is formatted into blocks, with provision made for synchronisation. There are two types of format, one for signals that occur in bursts, such as sound-in-syncs and B-MAC, and the other for continuous signal channels.

The basic operation of the decoder can be outlined with


Fig. 1: Dolby ADM decoder.
reference to Fig. 1. After demodulation the signal components, which consist of audio data for each channel at a typical bit rate of $200-300 \mathrm{~kb} / \mathrm{sec}$ and the control data at the half line rate of $7.8 \mathrm{~kb} / \mathrm{sec}$, are separated out using suitable filters. The audio data is clocked into a multiplier stage as a bipolar signal: the step-size data acts as the multiplying constant. The audio data is then converted to analogue form using a leaky integrator. De-emphasis control works in a similar way, but instead of using the control signal to vary the gain the amplifier stage involved operates as a variable, single-pole frequency de-emphasis network. The decoder is available in i.c. form (the Signetics NE5240), is simple and is relatively insensitive to component tolerances.

## MAC/Packet Systems

The overall frame structure for the C-MAC and D-MAC systems, with packet sound channels, is shown in Fig. 2. The only significant difference between the two systems lies in the carrier modulation method employed. C-MACl Packet uses 2-4 phase shift keying (a form of QPSK) modulation of the common sound and vision carrier while D-MAC/Packet uses duo-binary coding which can be amplitude or frequency modulated on to a separate sound carrier.

Each $64 \mu \mathrm{sec}$ line period contains 1,296 sampling points, which is equivalent to a sampling frequency of 20.25 MHz . The audio channels are sampled at 32 kHz , quantized into 14 bits per sample, and then coded in twos-complement form. For stereo the left and right channels are sampled simultaneously, coded separately and transmitted alternately. The sound and data bits are organised into 164 packets, each of 751 bits, in two sub-frames. We thus have a total of 123,164 bits which have to be transmitted in 40 msec , equivalent to a bit rate of $3.0791 \mathrm{Mb} / \mathrm{sec}$. This total capacity can be subdivided in many ways. Depending on the methods of coding and level of error protection employed, some of the possibilities include: three linear stereo sound channels with basic error protection; four companded stereo sound channels with basic error protection; two linear stereo sound channels with extended error protection; three companded stereo sound channels with extended error protection; or the equivalent in mono or dual-language channels.

The error protection systems used are as follows: (1) Linear mode, first level. One even parity bit is added to the eleven most significant bits of each sample. (2) Linear mode, second level. An extended Hamming code $(16,11)$ is applied to the eleven most significant bits. This is capable of correcting single-bit errors. (3) Companded mode, first level. One even parity bit is added to the first six most significant bits. (4) Companded mode, second level. An extended Hamming code $(11,6)$ is added to each sample. This will correct most single-bit errors.

The companding system is similar to that employed with NICAM-3 (see later), which is used for processing digital sound in studios etc. After sampling, the sound plus data burst is organised into blocks of 3214 -bit samples. These are then compressed to ten bits each, using a scaling factor determined by the magnitude of the largest sample in the


Fig. 2: Frame multiplex structure used for C-MAC and DMAC Packet systems (not to scale).
block. The scaling factor is encoded into the parity bits for each block to indicate the degree of compression. This scaling factor is extracted at the receiver using majoritydecision logic which also restores the original parity. The decoded scaling factor is then used to expand all the samples in the relevant block.

To minimise the effects of burst errors the 751 bits in each packet are interleaved. An energy dispersal or spectrum shaping technique is applied after interleaving, to randomise the data stream. This is done to all except the first seven bits in each line and the data in lines 624 and 625. The process consists of adding the output of a PRBS generator with a period of $2^{15}-1=32,767$ bits to the data stream by means of exclusive-or logic. The PRBS generator runs at 20.25 MHz and is initialised at the start of every frame so that the first addition always applies to bit eight of line one.

## C-MAC Modulation/demodulation

With the 2-4 PSK modulation system logic one is represented by a $+90^{\circ}$ phase shift while logic zero is represented by a $-90^{\circ}$ phase shift. There are three basic ways of demodulating such signals. If the carrier/noise ratio is high, typically greater than 16 dB , it's possible to use the vision f.m. discriminator to recover the audio/data signal as well. More commonly however either a coherent or a differential demodulator is used. A coherent demodulator detects the incoming signal and compares it with a highly stable reference signal: any instability leads to bit errors. Since the received signal is in the form of DPSK, differential demodulation can give better results: with the received data in the form of phase differences in successive intervals, these differences can easily be detected by comparing the received signal with itself after a delay of one bit period.

## D-MAC Modulation/demodulation

The bipolar duo-binary sound and data signal is in analogue form, with a bandwidth of only $10 \cdot 125 \mathrm{MHz}$, for a channel bit rate of $20.25 \mathrm{Mb} / \mathrm{sec}$. After demodulation the sound signal can be recovered by full-wave rectification followed by slicing at the half amplitude level.

## D2-MAC Sound Channel

The D2-MAC/Packet sound channel has the same format as D-MAC/Packet except that provision is made for only one sound and data sub-frame in the same approximately $10 \mu \mathrm{sec}$ period. The reduced bit rate of $10 \cdot 125 \mathrm{Mb} /$

(b)

D866
Fig. 3: The VIMCA system. (a) Block diagram of the encoding arrangement. (b) Decoder block diagram.
sec allows for one high-quality stereo channel plus a lower grade audio channel and a limited data service. The total bandwidth of the sound and vision channels is just under 13.5 MHz , allowing transmission over current cable networks.

## Compatibility of MAC Systems

In all the MAC variants described above the demodulated digital signal is processed in a manner complementary to the sequence used for encoding. That is, the signal is first descrambled to remove the PRBS energy dispersal component, de-interleaved, expanded from 10 to 14 bits and finally checked for errors. This common arrangement, plus the similarities in the sound and data frame multiplexes, means that universal chip sets that will automatically recognise and decode whatever system is in use are likely to be made available to setmakers.

## VIMCA System

An important point that has to be considered when planning to add stereo sound to an established mono TV network is the cost of modifying all the transmitters.

The Australian organisation IRT Ltd. has developed a bolt-on system that provides a neat solution. It's known by the initials VIMCAS (vertical interval multiple channel audio system) and can also be used with VCRs, again without modification being required. Basically, the system incorporates time-compressed and companded audio signals in spare line periods during the field blanking interval. Each line can accommodate an audio base bandwidth of approximately 4.7 kHz , so that six lines will provide a pair of stereo channels 14 kHz wide. Multiple lines can alternatively be used for dual-language or data transmissions.

Fig. 3 shows the general principles involved, (a) for encoding and (b) for decoding. We'll consider encoding first. The analogue audio signal in each channel is first band limited and compressed, then sampled, quantized and loaded into a digital memory. During the appropriate video line it's read out of the memory at a very much higher rate, thus achieving time compression. The signal is then
converted back to analogue form and is gated into the video signal. The bandwidth of the time-compressed audio signal is about 2.5 MHz , which is well within the capacity of the video channel. Decoding is done in a complementary manner, as shown in Fig. 3(b). Any additional channels require their own AD converters and digital memories but can share the DA converter.
When several contiguous lines are used for wideband audio there's signal duplication at the end and beginning of successive lines. The signal at the beginning of a line, where corruption by interference or distortion is most likely, can thus be discarded.
The system has been found to be very flexible in operation - it's possible to mix wide and narrow band signals without cross-talk. Scrambling can be provided while the signal is in digital form or simply by alternating the line sequences. When the system is used with a video tape recorder the signals are not affected by the head switching and, due to the method of synchronism, wow and flutter are said to be negligible.

## NICAM 728

The UK standard for terrestrially transmitted digital stereo TV sound channels is NICAM 728 . Let's briefly look at the history. The West German dual-carrier system was extensively tested in the UK, with the PAL system I standard. It was found to be almost impossible to include a second sound carrier between 6 and 8 MHz without causing unacceptable interference to either the vision or the primary sound carrier. With systems B and G, used elsewhere in Europe, the primary sound carrier is at 5.5 MHz with respect to the vision carrier, leaving enough spectrum space to avoid the interference problems found in the UK. Over the years BBC and IBA engineers have developed considerable expertise in digital processing of the TV sound channel - from the sound-in-syncs system used since the late sixties for sound links between studios and transmitters to the more recent work on MAC systems. Starting with this background BBC engineers developed the system that has come to be known as NICAM 728 NICAM relates to the companding system employed (near-instantaneous companded audio multiplex) while 728 indicates the digital data rate used.
NICAM 728 has a second subcarrier at a level of -20 dB relative to the peak vision carrier and spaced 6.552 MHz above it $(6.552 \mathrm{MHz}=9 \times 728 \mathrm{kHz})$. This carrier is differentially modulated by the digitally encoded signals for both channels of the stereo pair. The present 6 MHz f.m. sound channel is retained in the interests of compatibility with current mono receivers.

The digital subcarrier is quadrature (four phase) PSK modulated: each resting carrier phase represents two bits of data, thus halving the bandwidth required. Because of the differential encoding (DQPSK) only the phase changes have to be detected at the receiver, the bits to phase change relationships being as follows: $00=-0^{\circ}$ phase change; 01 $=-90^{\circ}$ phase change; $10=-270^{\circ}$ phase change; $11=$ $-180^{\circ}$ phase change.

Pre- and de-emphasis to CCITT recommendation J176.5 dB boost or cut at 800 Hz - is applied either while the sound signal is in analogue form or by means of digital filters while it's in digital form. The left and right channels a. : cimultaneously sampled at 32 kHz , then coded and quantized separately to 14 -bit resolution and transmitted alternately at a frame rate of 728 bits per millisecond ( $728 \mathrm{~kb} / \mathrm{sec}$ ).

The NICAM compander processes the 14-bit samples in


Fig. 4: Coding scheme for NICAM 728 companding.


Fig. 5: NICAM 728 frame multiplex.
the manner shown in Fig. 4. The rule for disregarding bits can be summarised as follows: the most significant bit (MSB) is retained and the four following bits are deleted when they are of the same consecutive state as the MSB; if this leaves a word of more than ten bits the excess bits are deleted from the region of the least significant bit (LSB). A single even parity bit is added to check the six most significant bits in each word. The data stream is then organised into blocks of 32 11-bit words in twos complement form.

The magnitude of the largest sample in each block is then used to determine a 3 -bit scaling factor, which is encoded into the parity bits for that block. A majority decision logic circuit is used in the receiver to extract the scale factor - this process also restores the original parity pattern.

Two blocks of data are then interleaved in a $16 \times 44$ (704 bits) matrix to minimise the effects of burst errors. Adjacent bits in the original data stream are now 16 bits apart.

A transmission frame multiplex of the form shown in Fig. 5 is then organised. Additional bits are used as follows: eight bits form a frame sync word (framing word); five control bits select the mode of operation ( $\mathrm{C} 0-\mathrm{C} 4$ ); eleven additional data bits are reserved for future developments. The modes are as follows: stereo signals consisting of alternate channel A and B samples; two independent mono signals transmitted in alternate frames; one mono signal plus one $352 \mathrm{~kb} / \mathrm{sec}$ data channel on alternate frames; one $704 \mathrm{~kb} / \mathrm{sec}$ data channel; other ideas not so far defined.
After the interleaving of the 704 sound data bits ( $64 \times$ 11-bit samples) the complete frame, except for the framing word, is scrambled to provide energy dispersal. This is


Fig. 6: Decoding the NICAM 728 stereo signal.


Fig. 7: DQPSK signal decoding.
done by adding via exclusive-or logic a PRBS of length $2^{\prime \prime}-$ 1. The PRBS generator is reset on receipt of the framing word.

To limit the bandwidth the data stream is passed through a spectrum shaping filter that removes much of the harmonic content of the data pulses. This, combined with the action of a similar filter in the receiver, produces an overall response that's described as having a full or 100 per cent cosine roll-off.

The data stream is finally divided into bit pairs to drive the 6.552 MHz subcarrier's DOPSK modulator.

## Decoding NICAM 728

NICAM 728 decoding is shown in block diagram form in Fig. 6. The secondary sound channel's subcarrier appears at either 32.948 MHz or 6.552 MHz depending on the arrangements used in the receiver's i.f. strip.

The spectrum shaping filter forms part of the system's overall pulse shaping and has an important effect on noise immunity. Overall filtering ensures that most of the pulse energy lies below a frequency of 364 kHz (half bit rate).

The QPSK decoder recovers the data stream which is scanned by the framing word detector so that the start of each frame is located in order to reset the PRBS generator. The PRB sequence is then added to the data via the exclusive-or gate to provide descrambling (energy dispersal signal removal). De-interleaving is also synchronised by the arrival of the framing word. Standard procedures are used for error control, which is carried out within an i.c. The operating mode detector searches for the control bits $\mathrm{C} 0-\mathrm{C} 4$ to set up the data and audio stage switches automatically, the data outputs being those for the 352 or $704 \mathrm{~kb} / \mathrm{sec}$ data channel options. The expansion circuit functions in a complementary manner to the compressor, but uses the scaling factor to expand the 10 -bit data words into 14 -bit samples. The data stream is finally converted back into analogue form for feeding to the audio amplifier stages. These should be designed to a very high standard the audio quality provided by NICAM 728 approaches that of the compact disc.

The DQPSK decoder is a particularly complex item
that's fortunately available in i.c. form - the block diagram shown in Fig. 7 is very much simplified. The two main sections are concerned with recovery of the carrier and the bit-rate clock. The first section employs a voltagecontrolled crystal oscillator running at 6.552 MHz and two phase detectors to regenerate the parallel bit pairs, which are referred to as the I and Q signals (in-phase and quadrature). A second similar circuit, locked to the bit rate of 728 kHz , is used to synchronise and recover the data stream. Parallel adaptive data slicers square up the data pulses and the DQPSK signals are then decoded by differential logic. The bit pairs are finally converted to serial form.

A practical decoder incorporates a further phase detector circuit driven from the $Q$ chain. This is used as an amplitude detector which generates a muting signal if the 6.552 MHz subcarrier is absent or fails. The audio system is then switched over to 6 MHz f.m. mono sound.

## Current Status of NICAM 728

Both the BBC and the IBA are currently involved in a transmitter replacement programme and plan to add NICAM 728. The BBC has announced that a regular service with NICAM 728 is unlikely to start before 1991, but the IBA has hinted that its services could start earlier. In the meantime, the Swedish and Hong Kong broadcast services have taken up the system and expect to be operational some time this year. As a result of the similarities with the MAC/Packet systems it's expected that chip sets for decoding will soon be available at a reasonable cost. Texas Instruments and Toshiba have both stated that they could have chips available at very short notice, while JVC has announced that it already has a TV receiver and VCR with digital stereo capability ready for launch as soon as the services come into operation.

## Correction

Finally, a correction to Part 2 last month. 2" - ' in the first line of the second column, page 271 , should have read $2^{n}-1$. As printed there would be only four PRBS states instead of the seven listed in Table 2.

# Service Bureau 

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## GRUNDIG CUC220 CHASSIS

After the set has been running for about twenty minutes the chopper transformer starts to buzz. This increases, with increased contrast or brightness. Eventually the set will go to standby intermittently. All the secondary supplies seem to be in order.

Replace R632 (100k $\Omega$ ), C631 ( $100 \mu \mathrm{~F}$ ) and R631 $(0 \cdot 68 \Omega)$. If the problem remains the TDA460) chopper control chip IC631 is suspect.

## FERGUSON 3V23

The picture is stable but there's very bad flutter on sound. This fault got worse over a period of six months and is now so bad that listening is unpleasant, especially with tapes recorded by the machine. Changing the main capstan drive belt has had no effect. I've bought a second machine to use while the first is put right and find that the same fault is beginning to occur with this one.

First check that the back-tension is correct and not varying - watch the back-tension lever in operation. If this is all right listen to the sound carefully. If the problem is amplitude variation (loud/soft) try zenith adjustment of the audio head - tilt its top slightly outwards. If the problem is frequency variation (wow) concentrate on the operation of the capstan. Clean the capstan and pinch roller - replace the latter if it's eccentric or binding. The capstan speed could be varying due to a faulty motor, but check the capstan servo circuit first, setting up as outlined in the manual - an oscilloscope is almost essential for this.

## ITT CVC9 CHASSIS

There are horizontal black lines at the top of the picture, in bands, decreasing in intensity from the top of the screen. Very occasionally these lines are not present. Various items in the line/field blanking circuits and the line output stage have been checked.


A common cause of this problem is ringing in the scan yoke (field section) or the pincushion distortion correction transductor. Check the following: R364, R362 (pincushion amplitude), L125 and R354 (vertical shift). A dry-joint on the transductor L121-4 or one of the above components is quite common.

## PHILIPS CTX-E CHASSIS WITH TELETEXT

There's a great disparity in the brightness levels between teletext and programme displays. When the brightness is set for a normal picture the teletext is blindingly bright. Are most sets sold like this? I gather there's no means of adjustment.

The amplitude of the teletext display is set by a small potentiometer which is situated between the SAA5050 character generator chip and plug V5. There are two potentiometers in this area on the teletext decoder board. The one nearest the edge adjusts for minimum judder in the mixed mode, the one that's farther in setting the brightness. These adjustments are not present on later boards. With these the teletext and picture brightness can be set separately by means of the remote control handset.

## FERGUSON 3V22

Considering its age this machine records and plays back quite well. The problem is that with prerecorded tapes the colour smears to the left. Adjusting the tracking control alters the condition for a second or so, then the colour returns to its original position.

This problem can be caused either by incorrect tape path alignment, which is common with later versions of the 3 V 22 , or by incorrect or varying drum or capstan speed. By adjusting the tracking control you are momentarily altering the drum speed so that the position of the heads with respect to the information on the tape alters, hence the fault clears momentarily. The action required to restore correct operation is to replace the belts, pinch roller and take-up clutch, then go through the full alignment procedure for the audio/servo board, taking particular care with the drum and capstan adjustments. If the problem persists, use a scope to check the pulses from the control and pickup heads. Finally, it may be necessary to carry out complete realignment of the tape path.

## ITT CVC1203 CHASSIS

When this set first came in we found that the mains bridge rectifier's reservoir capacitor C658 was leaky. Prior to this the set was reported to have "gone off" on a few occasions. No fault was noticed during a three day soak test, so as a precaution the chopper and line output transistors were replaced and the $h . t$. was checked. The set is now going off again, the first symptom being loss of station before the set reverts to the standby mode.

We suggest that you start by replacing R716 (150k $\Omega$ ) in the power supply and $\mathrm{C} 614(100 \mu \mathrm{~F})$ in the line generator circuit: these components are troublesome in this chassis. If necessary then suspect the 12 V regulator chip IC751 (type 7812) - after checking for dry-joints around the chopper and line output transformers.

## NEC PVC470E

After replacing the video heads the picture obtained is excellent. Record and playback of tapes recorded since the head change is normal but with tapes recorded before the head change there's a tracking problem which is usually confined to the lower part of the picture. The problem can be described as bright white blips with small tails. Moving
the tracking control either way from centre makes the problem worse, i.e. more blips with horizontal white lines. I suspect that the drum entry guide needs adjustment.

You are almost certainly right in suspecting that a tape path problem is present. Since the mistracking is mainly confined to the bottom of the picture however it's more likely that the problem lies with the exit guide. Before carrying out adjustment (consult the manual) clean the guide, head drum assembly etc. thoroughly and ensure that the head screws are tightened evenly.


303
Each month we provide an interesting case of
TV/video servicing to exercise your ingenuity. These are not trick questions but are based on actual practical faults.

This month's test case concerns an Hitachi VT8500 VCR, though the symptom and cure would apply equally to any VCR, whether of the VHS or Betamax format. It was a puzzler indeed. Since solving it however we've come across similar symptoms in other machines - and in every case the cause of the problem has been the same.

The fault was confined to the recording process: playback of a known good tape was perfect every time. When a tape recorded by the machine was played back the picture tended to roll and judder vertically - whether played back on the machine itself or another one. Vertical stability of the monitor's picture depends on the field sync pulses of course, so our first step was to hook an oscilloscope to the playback head amplifier.

The display on the scope's screen is shown at the top in Fig. 1. An unusual sight: the leading edge of every other r.f. envelope from the heads had a hole in it, wide enough to knock out some of the field sync pulse - which is almost the first thing to be recorded during each head sweep across the tape. The edges of the holes were quite steep and sharply defined, quite unlike the bottle-neck effect produced by a misaligned tape entry guide. Guide problems seemed unlikely anyway since the r.f. output envelope from one head was perfectly square and normal. Surely any


Playback R. E envelope

Drum flip-flop pulses
Fig. 1: Top, the waveform at the faulty machine's playback head amplifier. Bottom, the correctly timed 25 Hz drum flipflop waveform.
mechanical problems would effect both heads equally? We couldn't imagine any tape path or head faults that would give rise to this strange effect. So it seemed to be some sort of electrical fault.

When we monitored the luminance writing current during record we found that it was continuous and of the correct amplitude. Next we closely examined the 25 Hz drum flip-flop pulse waveform during record and playback. It was straight-sided, symmetrical and correctly timed, as shown at the bottom in Fig. 1. In desperation we phoned the owner and quizzed him closely on how and when the fault had developed. It seemed that the machine had been perfectly all right until it had been taken to a repair shop (not ours) with the complaint "woolly sound". Since being returned with a hefty bill the sound had been better but the picture had bounced and rolled like a ball. Our morale sank. The "fault" may have been the result of hamfisted twiddling, modification or bodgery . . .

In an attempt to analyse the fault symptom in greater detail we played back the tape on another, good machine and watched the playback r.f. envelope. The results were just the same. As soon as the section of tape recorded by the Hitachi machine changed to the machine's own recording the shape of the envelope returned to normal. We noticed a strange effect however, and this was the key to correct diagnosis. Two or three seconds before the end of each playback of a test recording session on the Hitachi machine the shape of the r.f. envelope returned to normal the hole had disappeared! This happened regardless of the machine used for playback. Suddenly we knew the answer! What was it? See next month.

## ANSWER TO TEST CASE 302 - page 291 last month -

The situation outlined last month arose from inexperience on the part of the technician sent to deal with the problem in the field. The very bright raster displayed by the 16in. ITT set couldn't have been caused by an increase in the c.r.t.'s cathode voltages - indeed the increase was a result of the technician's reaction and that of the beamcurrent limiter circuit to the high brightness fault, the technician backing off the manual brightness control while the beam-limiter circuit pulled down the contrast level, both in ineffectual attempts to restore a normal display.

As Sage quickly twigged, the key to the problem lay in the fact that adjustment of the tube's first anode voltage had no effect on the brightness of the raster. It should have done! His conclusion was that the first anode voltage was excessively high and was unaffected by the first anode potentiometer's setting. At no time had the field technician checked the voltage at pin 10 of the tube! The cause of the trouble lay in R46A of course: this resistor links the earthy side of the potentiometer to chassis, forming part of the potentual divider chain. Its body had cracked.

The value of R 46 A is $750 \mathrm{k} \Omega$, which is not normally carried as a spare - it's not a preferred value. Since the technician didn't have two $1.5 \mathrm{M} \Omega$ resistors to connect in parallel he fitted an $820 \mathrm{k} \Omega$ resistor and readjusted R47A to obtain the correct black level.

[^2]

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\hline 2SA－769 \& $\underline{1.50}$ \& 2SB－688 \& 51.25 \& 2SC－1403 \& 51.50 \& 2SC－3182 \& 92.20 \& AN－253 \& c0． 65 \& BA－333 \& 81.00 \& CA－3420AE E3．05 \& ${ }^{\text {CA }}$－3140E \& ${ }_{5} 0.50$ \& V－5480 \& ${ }_{5} 1.55$ \& \& B8－2020 \& 50.75 <br>
\hline 25A． 771 \& 51.50 \& 2SB－705 \& 52.50 \& 2SC－1413 \& 03.00 \& 2SC－3284 \& 51.50 \& AN－262 \& 51.10 \& 8A－340 \& 50.75 \& TIP－29A \& CA－3089 \& $\underline{51.15}$ \& V－7450 \& ${ }_{5} 1.30$ \& \& ${ }_{\text {BR }} \mathrm{BR}-2320$ \& ${ }_{c}^{50.75}$ <br>
\hline 2SA．794 \& c0．60 \& ${ }_{2 S B}$ ST16 \& ${ }^{50.30}$ \& 2SC－1445 \& 81.00 \& 2SC－3298 \& 51.50 \& AN－272 \& E． 9.90 \& BA－343 \& ${ }_{50} 50.75$ \& TIP－23A，B \& \& \& $\mathrm{V}-8600$
$\mathrm{~V}-5475$ \& ${ }_{51.45}$ \& ORTOFON \& 8R－2325 \& c0．75 <br>
\hline 2SA－798 \& 50.60 \& 2SB－717 \& ${ }^{1} 0.60$ \& ${ }^{2 S C-1446}$ \& $¢_{50.75}$ \& 2SC． 3506 \& $\underline{5.30}$ \& AN－301 \& ${ }_{5} 9.35$ \& BA－402
BA－527 \& c0． 50

50.97 \& TIP－3AA， 8 ¢0． 27 \& AKAS： \& \& V－5475 \& 1.45 \& WMEADSHELL \& CR－1620 \& c0．75 <br>
\hline 2SA－808 \& \＄1．50 \& 2SB－718 \& £0．75 \& 2SC－1447 \& 10.60 \& 2SC． 3519 \& 51.50 \& AN－302 \& $\underline{5} .50$ \& BA－527 \& ${ }_{5} \mathrm{E} .975$ \& TIP－30C \& Vs－10 \& £0．78 \& Ethguson \& \& CAATAIOGES \& CA－2025 \& c0． 75 <br>

\hline 2SA－817 \& $\underline{0.15}$ \& 2S8－757 \& ¢1．30 \& 2SC． 1454 \& $\ldots 3.50$ \& $2 \mathrm{2C}-8050$ \& c0． 18 \& AN－303 \& $\mathrm{ESP}^{20}$ \& BA－536 \& | 51.45 |
| :--- |
| 50.85 |
| 8.850 | \& TIP－31 \& VS－2EG／5EG \& ${ }^{20.78}$ \& 3 V 00 \& $\underline{2.55}$ \& $\begin{array}{ll}\text { VMS－3U } & 77.50 \\ \text { VMS．3S }\end{array}$ \& CA－2032 \& ¢0．75 <br>

\hline 254.844 \& ¢0．10 \& 2SB－772 \& ${ }^{50.50}$ \& 2SC－1509 \& ${ }_{50} 50.45$ \& 2S0－198 \& 51.90 \& AN－315
AN－ 318 \& ${ }_{51.00} 85$ \& BA－612 \& ${ }^{50.85}$ \& TIP－31A．B ¢ $¢ 0.22$ \& VS－7300 \& £1．35 \& 3V16 \& 81.95 \& vMs．3S $\quad 27.50$ \& CA－2316 \& co． 75 <br>
\hline 2SA－850 \& $\mathfrak{5 0 . 3 0}$ \& 2SB－837 \& 10.50 \& 2SC－1567 \& $\underline{0.50}$ \& 2SD－200 \& 53.10 \& AN－318 \& ¢5．75 \& BA－714 \& $\underline{50.30}$ \&  \& \& $\$ 1.60$ \& 3V22 \& 92.00 \& \& CA－2420 \& ¢0．75 <br>
\hline 2SA－893 \& ¢0．30 \& 2SB－857 \& 50.50 \& 2SC－1568 \& 20．45 \& 2SD－235 \& 50.35 \& ${ }^{\text {A }} \mathrm{N}$－340 \& ¢1．20 \& BA－1310 \& $\mathcal{E c}^{1} .65$ \& TIP－32 \& \& \& ${ }^{3} \mathrm{~V} 23$ \& $\mathrm{c}_{50.75}$ \& \& CR－2430 \& c0． 75 <br>
\hline 25A－896 \& ¢0．35 \& 2SC－352 \& 50.60 \& 2SC－1577 \& ¢7．70 \& 2SD－288 \& ${ }^{5} 0.75$ \& AN－360 \& ¢0．75 \& BA－5102 \& ${ }_{51.20}$ \&  \& FISHER \& \& 3V29 \& ¢0．75 \& CALCULATOR \& \& <br>
\hline 2 SA－916 \& $¢_{00.18}$ \& 2SC－372 \& ${ }^{50} 0.10$ \& 2SC－1550 \& c0． 50 \& 250.299
$2 \mathrm{SD} \cdot 313$ \& 51.50
50.30 \& ${ }_{\text {AN }}^{\text {AN－51010 }}$ \& ${ }_{57.50}^{50}$ \& BA－5402 \& ¢1．
¢1． 20 \&  \& VBS－7000 \& 52.40 \& DRIVE \& \& MiCPO \& Celis） \& Round <br>
\hline 2SA－921 \& $\underline{50.10}$ \& 2SC－380 \& ${ }^{0} 0.12$ \& 2SC． 51514 \& ${ }^{2} 0.75$ \& 2SD－313 \& 50.30
50 \& AN－5111 \& $\xi_{5.50}$ \& BA－5404 \& ¢1． 20 \& TIP－33A $\quad 5050$ \& VBS－9000 \& $\underline{50.80}$ \& \& \& batteaies \& 810 （N） \& c0． 42 <br>
\hline 2SA．940 \& ¢0．45 \& 2SC－458 \& c0． 15 \& ${ }^{2 S C}$ C－1584 \& ¢5．50 \& 2SD－315
$2 \mathrm{SO}-325$ \& ${ }^{2} 0.75$ \& ${ }_{\text {AN }}{ }_{\text {AN }}$－5410 \&  \& BA－6109
HA－1124 \& ¢1．40 \& TIP－41 ¢0 22 \& \& \& RECOAOE \& \& RW－40 50.48 \& 813 （0） \& ¢0． 48 <br>
\hline 254．950 \& ¢0．25 \& 2SC－460 \& $\mathrm{E}^{0} .06$ \& 2SC－1586 \& ${ }^{2} 5.50$ \& $250-325$
$250-352$ \& ${ }_{50}{ }^{50.45}$ \& AN－5431 \& $\sum_{51.80}$ \& HA－1124 \& ¢1．25 \& TIP－41B，C［ ¢ 23 \& HITACH \& \& TURNTAB \& \& RW－42 \& 814 （C） \& ${ }^{\text {c0．}}$ ． 38 <br>
\hline 2S4．958 \& ¢0．75 \& 2SC－495 \& 50.60 \& 2SC－1627 \& 50.20 \& $250-352$
$250-357$ \& ¢0．50 \& AN－5435 \& ¢1．80 \&  \& 81.25 \& IIP－42 ¢0 25 \& VT－5000E \& \& \& \& RW－44 \& 815 （AA） \& ¢0．20 <br>
\hline 254－968 \& ¢0．75 \& 2SC－496 \& ¢0．75 \& 2SC－1667 \& 51.40
80.75 \& $250-357$
$250-358$ \&  \& AN－5440 \& \％2．15 \&  \& ${ }_{81.25}$ \& TIP－42A，B ¢0．22 \& JVC \& \& squar \& \& RW－47 ${ }^{\text {RW0．25 }}$ \& 824 （AAA） \& £0． 25 <br>
\hline 2SA－985 \& ${ }^{5} 0.60$ \& 2SC－497 \& ¢1．50 \& 2SC． 1669 \& 180.75
80.75 \& 2SD－358
$2 \mathrm{SD}-381$ \& ¢0． 35
c0．90 \& AN－5510 \& $\underline{\xi 2.50}$ \& HA－1151
HA－1156 \& 17.25
¢1． 30 \& TIP－42C \& HR－3330 \& $\underline{5200}$ \& 68×12 to \& ¢0．12 \& $\begin{array}{ll}\text { RW－48 } \\ \text { RW－49 } & \\ \text { ¢0．42 }\end{array}$ \& A1604（6L） \& ${ }^{22}$ <br>
\hline 2SA－992 \& c0． 30 \& 2SC－536 \& ${ }^{\text {colo }}$ \& 2SC－1670
2S． 1675 \& 50.75

50.19 \& 2SD－381 \& | c0． |
| :--- |
| c0． 75 | \& AN－5612 \& $\underline{¢ 2.80}$ \& HA－1196 \& E1．30 \& TIP－48 \& HR－7200 \& $\underline{50.75}$ \& 86x12

$120 \times 1.25$ \& \& | RW－49 |  |
| :--- | :--- |
| RW－410 |  |
| 50.45 |  |
| 0.45 |  | \& \& <br>

\hline 2SA－1048
2SA－1060 \& E1． 1.50 \& 2SC－644 \& 180.25
$\mathbf{5 1 . 9 5}$ \& 2SC． 17275 \& 50.18

c0． 50 \& ${ }_{2}{ }_{2}^{250-388}$ \& | ¢00． |
| :--- |
| 1.80 | \& AN－5700 \& ¢0．60 \& HA－1197 \& \＄1． 20 \& TIP－102 \& HR－3360 \& 51.95 \& $135 \times 1.25$ \& $\mathrm{c}_{0} 0.12$ \& RW－411 $£$ \& PHOTO \& <br>

\hline 2SA－1062 \& 51.20 \& 2SC－693 \& c0．25 \& 2SC－1756 \& 50.45 \& 2S0－389 \& $¢ 0.95$ \& AN－5722 \& ¢1．35 \& HA－1319 \& $\underline{51.45}$ \& TIP－105 \& HR－4100 \& 51.95 \& \& \& RW－413 ${ }^{\text {P0，45 }}$ \& Batteries \& <br>
\hline 2SA－1094 \& \＄1．90 \& 2SC－710 \& 50.20 \& 2SC－1760 \& c0．75 \& 2SD－400 \& $\underline{50.15}$ \& AN－5730 \& $\underline{51.35}$ \& HA－1366W \& 51.75 \& TIP－125 \& HR－6500 \& 92.25 \& FLAT \& \& $\begin{array}{ll}\text { RW－415 } \\ \text { RW－418 } & \text { ¢0．45 } \\ \text { ¢0．91 }\end{array}$ \& 887 （J） \& ${ }^{51.54}$ <br>
\hline 2SA－1102 \& 51.90 \& 2SC－717 \& $\mathrm{c}_{0} .25$ \& 2SC－1775 \& ¢0．15 \& 2SO－401 \& $\underline{50.45}$ \& AN－5732 \& £1．25 \& HA－1366WR \& 51.75 \& $\begin{array}{lll}\text { TiP．} 126 & 50.40 \\ \text { ¢0．40 }\end{array}$ \& HR－3300 \& $\underline{925}$ \& \& 0.25 \& RW－30 ${ }^{\text {ROP }}$ \& RPX－14 \& £1．45 <br>
\hline 2SA－1104 \& $\underline{\$ 2.05}$ \& 2SC－733 \& $\underline{0} 0.25$ \& 2SC－1815 \& £0．15 \& 2SD－426 \& £1．50 \& AN． 5738 \& c1．00 \& HA． 1367 \& ${ }^{\text {E1．}}$ ¢0 \& HCF4001BE © 0.18 \& HR－7700 \& E0．77 \& $88 \times 05 \times 4$ \& \& RW． $33 \quad 80.45$ \& APX－23 \& ¢1．23 <br>
\hline 2SA－1106 \& 51.50 \& 2SC－738 \& 50.25 \& 2SC． 1819 \& £0． 71 \& 2SD－428 \& 51.50 \& AN－5900 \& ${ }^{\text {c1．}} 50$ \& HA－1374 \& ¢1．99 \& HCF4008BE 00.50 \& HR．7650 \& ¢0．77 \& $88 \times 05 \times 5$ \& \&  \& RPX－27 \& 9.05 <br>
\hline 2SA－1110 \& $\underline{1} .45$ \& 2SC－741 \& 51.95 \& 2SC－1845 \& £0． 15 \& 2SD－438 \& ¢0． 30 \& AN－6248 \& ¢1．20 \& HA 1337 \& 52.00 \& HCF4017BE £0．52 \& HADOMAL \& \& $122 \times 0.5 \times 5$ \& \& RW－37 50.31 \& \& $\sum_{50.35}$ <br>
\hline 2SA－1142 \& $\underline{\square} .90$ \& 2SC－783 \& 51.16 \& 2SC－1875 \& $\ldots 2.40$ \& 2SD－468 \& ¢0． 25 \& AN－6249 \& 11.20 \& HA－11225 \& ¢1．70 \& HCF4025BE ¢0．25 \& NV． 333 \& ¢1． 35 \& $189 \times 0.5 \times$ \& ${ }^{50.60}$ \& $\begin{array}{ll}\text { RW．} 39 & 50.52 \\ \text { RW－300 } & \\ 50.58\end{array}$ \& RPXX－625 \&  <br>
\hline 2SA－1145 \& 50.20 \& 2SC－789 \& 50.35 \& 2SC． 1890 \& c0． 20 \& 2SO－476 \& c0．45 \& AN－6250 \& ${ }^{50} .40$ \& HA－11227 \& 81.00 \& HCF4028BE £0．48 \& NV－8600 \& $\underline{81.65}$ \& $195 \times 05$
$205 \times 0.5$ \& 50.60
50.60 \& $\begin{array}{ll}\text { RW－300 } & \text { c0．} \\ \text { RW－310 } \\ 50.38\end{array}$ \& RPX－675 APX－825 \& <br>
\hline 2SA－1147 \& 51.90 \& 2SC－790 \& 20.96 \& 2SC－1906 \& $\mathrm{c}^{0} 0.25$ \& 2SD－478 \& ¢0．90 \& AN－6320 \& 52.00 \& HA．11235 \& 1.70 \& HCF4050BE E0．32 \& NV．777 \& $\underline{0.95}$ \& \& \& $\begin{array}{cc}\text { RW－310 } & \text { co．} \\ \text { RW－311 } \\ \text { E0．39 }\end{array}$ \& RPX－825 \& ${ }^{〔} 0.55$ <br>
\hline 2SA－1156 \& ${ }^{5} 0.60$ \& 2SC－828 \& ${ }_{5} 0.15$ \& 2SC－1913 \& ¢0．90 \& 2SD－525 \& $\underline{50.75}$ \& AN－6338 \& ${ }^{2} 5.00$ \& HA－11244 \& 81.65 \& HCF40103BE 5099 \& NV－7200 \& c0． 84 \& Cassemt \& ADS \& RW－313 \& TONGL \& <br>
\hline 2SA－1180 \& ¢1．80 \& 2SC－829 \& ¢0．15 \& 2SC． 1914 \& E0．15 \& 2S0－526 \& $\underline{1} 0.75$ \& AN－6341 \& $\underline{52.80}$ \& HA－11251 \& ${ }^{5} 0.80$ \& HCF40106BE CO． 35 \& NV－7000 \& c0．95 \& \& \& RW－315 $\quad$ E0．42 \& （Supe \& <br>
\hline 2SA－1220 \& ${ }^{2} 0.45$ \& 2SC－839 \& $\underline{0} 2.25$ \& ${ }^{2 S C}$－1922 \& $\underline{\mathrm{E}} .50$ \& 2SO－600 \& ${ }^{60.90}$ \& AN－6342 \& ${ }^{¢ 1} .50$ \& HA－11423 \& $⿳ ⺈ ⿴ 囗 十 大$ \& L－123CTB 81.30 \& NV．600 \& ¢1．45 \& STERED \& $¢ 1.50$ \& RW－316 \& $\mathrm{AC}^{\mathbf{3}}$（PP） \& ¢0．52 <br>
\hline 2SA－1232 \& ¢1．80 \& 2SC－929 \& $\underline{50.15}$ \& 2SC－1941 \& c0． 40 \& 2SO－612 \& ${ }^{50.40}$ \& AN－6360 \& 92.80 \& HA－12902 \& ¢1．70 \& \& \& \& \& \& \& \& <br>
\hline 2SA－1262 \& $\underline{1.55}$ \& 2SC－930 \& ${ }^{2} 0.15$ \& 2SC－1942 \& $¢ 2.70$ \& 2S0－613 \& ${ }^{2} 0.65$ \& AN－6551 \& ¢1．00 \& HA－12017 \& ¢1．30 \& \& EASE \& \& FO \& \& LISTEO AB \& \& <br>
\hline 2SA－1265 \& ${ }^{1} 1.30$ \& 2SC－941 \& ${ }^{\text {co }}$ 0．25 \& 2SC－1986 \& $\mathrm{c}_{50.45}$ \& 2SD－669 \& ¢0．45 \& AN－6651 \& ¢0．45 \& HA－12413 \& ¢1．30 \& \& ITEMS \& LIV \& Y IS S \& C \& 0 AVAILABI \& \& <br>
\hline 2SA－1303 \& ${ }^{\text {¢1．}}$［00 \& 2SC－945 \& ¢0．15 \& 2SC－2003 \& ${ }^{£ 0.25}$ \& ${ }_{\text {2SO－716 }}$ \& £0．85 \& AN－6884 \& \& HA－12411 \& c1．60
50 \& ABOVE P \& CES AR \& EX－V \& PRICE \& AN \& HANGE WITH \& UT NO \& <br>
\hline 2SB－324 \& $\mathrm{C}_{5} 0.45$ \& 2SC－959 \& ¢0．60 \& 2SC－2022 \& £0．30 \& 2S0－718 \& $£ 1.25$
El .30 \& AN－6912 \& ¢1．
$\mathbf{1 1 . 2 5}$ \& LA－1201 \& 60.75
81.60 \& PECIAL 0 \& DTATION \& ARE \& IVEN FO \& ARG \& AND EXPOP \& QUAN \& <br>
\hline 2SB－337
2SB－407 \& ¢1．
［1．30 \& 2SC－998
2SC－1012 \& E0． 60
c0． 80 \& 2SC－2073
2SC．2120 \& ${ }_{50.75}$ \& 2SO－733
2S0．745 \& $\underline{E 2.30}$ \& AN－7060 \& 11.25
81.60 \& LA－1365 \& ع1．20 \& FULL \& LISt AVA \& BL \& WITH O \& R 0 \& SAE PLEASE \& $\times 4^{\prime \prime}$ ． \& <br>
\hline 2SB－492 \& 20． 30 \& 2SC－1018 \& E0．75 \& 2SC－2229 \& ¢0．25 \& 2S0．748 \& 81.50 \& AN－7110 \& £1．20 \& La－3161 \& £1．20 \& \& ALL TH \& 00 \& ARE \& AN \& TOP QUALI \& \& <br>
\hline 2SB－507 \& c0．90 \& 2SC－1030 \& $\underline{\square} 2.20$ \& 2SC－2236 \& ¢0．18 \& 2S0－761 \& ¢0．45 \& AN－7116 \& ¢0．90 \& LA－3210 \& 50.45 \& ORDERS \& BELOW \& ． 00 \& EX－VAT） \& D P8 \& £0．78（For \& K．only \& <br>
\hline 2SE－511 \& c0． 90 \& 2SC－1050 \& ¢3． 20 \& 2SC－2240 \& £0．15 \& 2S0－8228 \& £4．50 \& AN－7117 \& ¢0．80 \& LA－3220 \& 51.00 \& BUT OR \& DERS AB \& VE $¢ 5$ \& 00 （EX－ \& ）$P$ \& P FREE（For U \& K．on \& <br>
\hline 2SB－512 \& ¢1．25 \& 2SC－1060 \& $\underline{50.45}$ \& 2SC－2274 \& $\underline{50.20}$ \& 2S0－837 \& $\underline{50.85}$ \& AN－7118 \& ${ }^{51.30}$ \& LA－3365 \& ${ }_{5}^{51.20}$ \& VISITIN \& G TIME： 1 \& AM $T$ \& 6PM（M \& N－FRI \& 10AM TO 12 \& M SAT． \& <br>
\hline 2SB－514 \& c0．49 \& 2SC－1061 \& ¢0．75 \& 2SC－2278 \& ¢0．75 \& $2 \mathrm{SD}-838 \mathrm{~L}$ \& 187.50
8175 \& AN－7130 AN－7140 \& ${ }_{5}^{50.60}$ \& LA－4100 \& 50.85
80.60 \& \& \& \& \& \& \& \& <br>
\hline 2SB－536 \& 20．50 \& 2SC－1115 \& $\underline{2} 290$ \& 2SC－2335 \& $\underline{\$ 1.10}$ \& 2S0－859 \& ${ }^{2} 0.95$ \& AN－7145 \& 27.20 \& LA． 4110 \& 51.20 \& \& \& \& \& \& \& \& <br>
\hline 2SE－537 \& ${ }^{2} 0.60$ \& 2SC－1116 \& $\underline{52.90}$ \& 2SC－2371 \& E0．50 \& 2SD－869 \& E3． 20 \& AN－7146 \& 12.28 \& LA－4112 \& 51.20 \& HLO \& TER \& \& Th \& \& \& \& <br>
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| 9600 | £25.00 | £20.00 | G11 Text | £50.00 | £35.00 |
| TX10 | £35.00 | £25.00 | Decca 80/100 | £15.00 | £10.00 |
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