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All correspondence regarding advertisements should be addressed to the Advertisement Manager, "Television", King's Reach Tower, Stamford Street, London SE1 9LS. Editorial correspondence should be addressed to "Television", IPC Magazines Ltd., King's Reach Tower, Stamford Street, London SE1 9LS.

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## QUERIES

We regret that we cannot answer technical queries over the telephone nor supply service sheets. We will endeavour to assist readers who have queries relating to articles published in Television, but we cannot offer advice on modifications to our published designs nor comment on alternative ways of using them. All correspondents expecting a reply should enciose a stamped addressed envelope. Requests for advice on dealing with servicing problems should be directed to our Queries Service. For details see our regular feature "Service Bureau". Send to the address given above (see "correspondence").

## this month

## 241 Leader

242 Long-distance Television
Roger Bunney
Reports on DX reception and conditions and news from abroad
247 The CED Video Disc System
Derek Snelling
The RCA CED video disc system was the outcome of heavy research and development expenditure over many years. It's an elegant solution to the problem of how to store video signals on a disc, doing so by means of frequency modulated capacitance variations. A look at the techniques involved and the players released in the UK late last year
250 Teletopics
News, comment and developments.
252 Vintage TV: The Ferguson 841T
Vivian Capel
The UK's leading setmaker didn't enter the TV field at the very start. When they did, shortly after the war, this is what they came up with.
253 VCR Clinic
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## 256 Letters

257 Next Month in Television
258 Servicing the Sony KV1400UB
David Botto
Another popular small-screen Sony colour set. The
interrelationships between the switch-mode power supply and the line timebase mean that a systematic approach is called for in dealing with the dead set symptom. This and other fault conditions you may come up against.
260 TV Fault Finding
Notes on TV faults from Mick Dutton, P. Hardy, M. S.
Barakat and John Coombes.
262 VCR Servicing, Part 26
Mike Phelan
The Ferguson $3 \mathrm{~V} 24^{\prime} \mathrm{s}$ drum servo system.
265 The Card Game is Over
Les Lawry-Johns
Monthly report from that famed servicing centre.
DX Signal Detector/Alarm
G. R. Exeter

Two PLL i.c.s are used as a narrow-band filter to detect the presence of the signal's line sync component, producing an output to gate the video signal through to a simple audio amplifier to provide an audible indication.
269 A Question of Black Level
Malcolm Burrel/
A black-level signal is useful for servicing, for editing
video programmes and other purposes. A simple black-
level generator is described.
270 Service Bureau
271 Test Case 255

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## Investment decisions

The Tele-Jector story is an intriguing one. It seems that the directors of a firm called London and Liverpool Trust, whose main operation is the distribution of business equipment (photocopiers, computers, etc.), got this bright idea for making money. A subsidiary, Tele-Jector, would instal projection TV equipment in pubs and clubs and would generate a substantial revenue from hire agreements. The clever bit is that it wouldn't cost anyone anything. The publican would be able to recover the cost of his Tele-Jector through the advertising revenue it would produce, and with enough TeleJectors out on lease the whole exercise would soon pay for itself so far as London and Liverpool Trust were concerned. Only it didn't work out like that.

The height of its fame came when Tele-Jector made an $£ 8 \mathrm{~m}$ bid, just over a year ago, to obtain exclusive rights to show Football League matches on its screens. London and Liverpool shares shot up to a high of 350 p . Unfortunately for Tele-Jector the bid fell through - the broadcasting authorities made a successful counterbid, and it transpired that not all League clubs would be prepared to participate. Even if they had, it would have taken some effort to recover the $£ 8 \mathrm{~m}$ and the cost of the other programme material provided in cassette form and the Tele-Jectors themselves. Looked at in this light the scheme looks rather less clever. It seems even less clever when you consider the underlying assumption - that pubs and clubs would be filled with avid Tele-Jector viewers. One wonders whether the directors of London and Liverpool Trust ever set foot in their local. If they had, they might have discovered that the box, when present, is seldom given much serious attention. At least the publicans appreciated that, and far fewer than were estimated to be necessary to make the venture pay decided to take it on.

At the end of the day London and Liverpool Trust were faced with the cost of providing programme material for an uneconomically small number of Tele-Jectors, and those publicans who had taken it on found that there was negligible advertising revenue. There was also some criticism of the leasing terms. London and Liverpool Trust's shares dropped to a low of 11 p , and liabilities of 18 m incurred by the project threatened the company's existence. A sad case of things going badly awry.

Is there a lesson to be learnt from this? Well, it could be an example of one of Parkinson's laws at work - that a group of people, in this case the directors, will often take a collective decision that each individually considers to be highly doubtful. It could also be an example of making an investment with a view to generating a quick profit. Of more relevance to television, doesn't it seem analogous to some of the headier notions of bonanzas to be made from developments such as cable and satellite TV? There was indeed a time when ITV was "a licence to print money", but it looks increasingly as if that was a once only state of affairs. The more TV there is, of one sort or another, the greater the competition and the harder the economic climate.

There has been much criticism that those responsible for investment in the UK seek too quick a return on their money. I'm not at all convinced that this is fair. In a highly uncertain world, is it reasonable to lock up funds in long-term projects whose outcome is difficult to foresee? We need industrial investment for sure, but investment in projects that don't turn out is worse than no investment at all.

In an interesting article in Electrical and Radio Trading, Erik Arnold argues that, amongst other things, the UK TV industry has reached its present sorry state due to lack of investment in new product technology. Those of us with fairly long memories might question this. For a start, successive governments through the fifties to the mid-seventies engaged in stop-go economic policies. Consumer taxes and credit control provided a simple means of putting such policies into effect, but meant that industries such as TV bore the brunt of the accelerator/brake business. This sort of thing hardly encouraged long-term planning - you can't be expected to undertake expensive investment when wondering how to finance a warehouse full of unsold TV sets awaiting the next "go". Firms nevertheless did invest in research and development, but if I read him right Erik Arnold argues that they didn't go about this in the right way. They left much of the technical research to component suppliers whilst concentrating on assembly. So we don't have integrated firms that can develop and produce their own tubes, i.c.s and systems through to the end product offered to the consumer.
This is largely due to the structure of the industry and is not necessarily a bad thing. It would hardly be sensible for every TV firm to produce its own types of tube - a case of reinventing the wheel time and again - while i.c. development is best left to those with semiconductor technology know-how. This doesn't mean that setmakers simply sit back and make use of whatever's on offer. That way does lead to disaster, in terms of unreliable consumer products. Developing reliable components that will survive in the stressful conditions of a TV set and the knock about home environment is a two way business between the component manufacturer and his customer, while UK setmakers have often taken the initiative in going to component manufacturers with their requirements and jointly working on the solution. Whether it would be better for component manufacturers and setmakers to be part of the same company is something that could be argued about endlessly. Provided they both have technical compentence it shouldn't really matter.

This does leave unanswered the criticism that the UK's industry has failed to come up with new product technology such as the VCR. For this you need a certain vision and confidence that perhaps we lack today. But really successful new products are not everyday occurrences. There is probably no harm in manufacturing under licence rather than undertaking the original research.

Please drop us a line to let us know your new address.

## FRONT COVER

Our thanks to Hitachi Sales (UK) Ltd. who provided this month's cover photograph showing the Hitachi Model VIP101P CED video disc player.

# Long-distance Television 

Roger Bunney

Now that we're into the new year it's possible to summarise the conditions during 1983. They were pretty good! Sporadic E signal propagation continued after its usual ending in late August, and there was a little F2/TE reception as the sunspot cycle dwindled. More encouraging were the good tropospheric openings towards the end of the year - indeed December both started and ended with spells of enhanced tropospheric propagation. There were good tropospheric openings during all months from late August - my own most interesting catch was ORF-1 (Austria) ch. E9 on September 26th, a 20 kW transmitter at Bruck near the Hungarian border.

Reports from our Australian friends suggest that they are enjoying a good SpE season, and one must hope that this heralds a good season ahead for us in the UK. With the 405 -line transmitter closures, many will be able to enjoy Band I/III relatively clear of local transmissions. This could be the last year of Band I being free from interference, so make the most of it!

What with other activities during December, reports of reception normally slacken off. From the letters received however it seems that the main event was the good tropospheric opening on December 27-29th. Reception was mainly from the E./S.E., with signals from W./E. Germany, the Benelux countries and France. The prevailing high-pressure system first gave a signal lift on the 27th, with strong French u.h.f. signals and W. Germany (just). There was further improvement in Band III and at u.h.f. on the 28 th, with a peak on the 29 th when signals extended from W. Germany across the UK as far as Wales. Unusually, the emphasis was on W. German Band V signals, reaching as high as ch. E60. The in-vision teletext/scrambled information from Paris ch. 8 was present, but not the mystery ch. 5. RTL (Luxembourg) was widely seen, the PM5534 test pattern carrying the identification "RTL PLUS". We've a report that RTL have been using both PAL and SECAM at their ch. E7 outlet. ATV activity was high during the period. F3YX, to the S.W. of Paris, was seen here at Romsey - a new station for me and a most welcome bonus to see out 1983!

Meteor shower/scatter reception has been remarkably
good, with two prime showers during the period. The Geminids produced active conditions over the $10-14$ th, the peak on the 13 th producing signal pings and more sustained bursts throughout Band I, reaching into the lower end of Band III (chs. E5, E6 and R5) on occasions. In early January the Quadrantids again produced MS reception - the peak appeared to be on the evening of the 3rd.

There was little SpE reception over the period. The $\log$ is as follows:
23/12/83 NRK (Norway) ch. E3.
30/12/83 Unidentified signals on chs. R1 and R2.
3/1/84 RAI (Italy) IA; unidentified signals on E3, 4.
4/1/84 RTVE (Spain) E2, 3, 4; RTP (Portugal) E2.
My thanks to Simon Hamer (Powys), Hugh Cocks (E. Sussex), Graeme Wilson (Cleveland), John Tellick (Surrey) and Ryn Muntjewerff (Holland) for reception reports.

Robert Copeman reports that December started off quietly in Australia so far as SpE reception is concerned, though conditions had opened up by the end of the month - with little evidence of double-hop propagation unfortunately. He mentions that the increasing number of Band II f.m. radio stations in operation there is giving greater scope for v.h.f. DXing. The Network $0 / 28$ (cultural service) is to loose most of its ch. 0 outlets with a move to u.h.f. operation only. This should ease DX reception by 1985.

Jim Maden reports a "lousy" season in S. Africa, with no sign of ZTV Gwelo ch. E2. This is most unusual and he wonders whether the transmitter is now off the air. Bulawayo ch. E3 has moved to Band III - due to interference from Europe! Jim views TSS-1 via the Ekran 714 MHz downlink, and comments on the flashing diamond caption previously mentioned - at TSS programme closedown. The translation of the Ekran caption reads "don't forget to switch off the television", with the words in red on a white diamond with a blue background. The centre words that flash are "switch off", and there's also an 800 Hz tone. During various October/November TE openings Jim received RTVE, RAI and RTP, the highest frequency signal being ch. E3/IA vision. Finally Jim mentions that TSS now use an African lady who speaks fluent Russian for the discussion of African affairs - so watch out for her on ch. R1 during the next SpE season. . .

## Radio Receivers

In the January column I mentioned the Lowe Electronics AR2001 scanner, which covers $25-550 \mathrm{MHz}$. Revco Electronics have since told us that the professional version of their SX200n will shortly be available at some $£ 400$. It


Left: The new Rumanian test pattern (similar to DFF/GDR), received by Ryn Muntjewerff from Bucharest on ch. R2. Centre: ATV reception in Holland by Ryn Muntjewerff, at 435 MHz . G3RJM card (Newcastle) with G8PZF background. Right: Tropospheric reception by John Tellick (Surbiton) from Anderlues, Belgium, during the September opening.
covers $26-520 \mathrm{MHz}$, with the only gap being $88-108 \mathrm{MHz}$. A.M. and f.m. signals are catered for: the sensitivity is $0.5 /$ $1 \mu \mathrm{~V}$ (f.m./a.m.) at v.h.f., $0.5 / 2 \mu \mathrm{~V}$ (f.m./a.m.) at u.h.f. The image rejection is 50 dB at v.h.f., 40 dB at u.h.f., with sharp selectivity. The unit will interface with certain NEC computers for outboard control, for example to increase the channel memory and for high-speed reprogramming. Perhaps more interesting is the availability of a prescaler that gives operation at up to 3 GHz ! More details can be obtained from Garex Electronics, 7 Norvic Road, Marsworth, Tring, Herts HP23 4LS (0296 668684).

Perhaps more down to earth is a range of Tandy (USA) portables for TV.sound. Their 1984 catalogue features four receivers, one hand held, with full system M v.h.f. channel coverage - one receiver has full u.h.f. coverage as well. The Portavision 55 covers low and high bands (Bands I/III) and has a 70 channel u.h.f. tuner (click stops plus fine tuning). The price is $\$ 79.95$. The Portavision 5 has similar v.h.f. TV coverage plus air/PMR at 108175 MHz and is perhaps the best for general v.h.f. coverage at $\$ 59.95$. The Personal Portavision covers just the low/high TV bands at $\$ 44 \cdot 95$. All three models include MW and $88-108 \mathrm{MHz}$, give good sound reproduction and are sensitive (there's a tuned r.f. stage for each band).

I'm using the Portavision 5, which I've realigned from ch. A2 sound to ch. E2, i.e. from 59.75 MHz to 53.75 MHz . By shifting the coverage in this way a gap has appeared at the top end of the low band, but ch. R3 $(83.75 \mathrm{MHz})$ can still be received. Apart from this gap there's full coverage over $53-220 \mathrm{MHz}$. The radio has proved to be very useful during SpE openings since it will resolve both f.m. sound and the a.m. vision buzz. In addition to the integral whip aerial, there's a Motorola (car radio) socket for an external aerial. You can't obtain these sets from Tandy UK (in part due to the battery/ 115 V a.c. operation) but you can obtain them by mail order from Tandy Corporation, Export Sales, Fort Worth, Texas 76102, USA. Note that you may get involved with import duty/VAT - it may be easier to obtain sets secondhand via a friend.

## News Items

Belgium: The French language ECS downlink channel "TV5" (uplink at a French station near Troyes) came into operation on January 2nd, with programmes from 19002200 nightly. TDF-A 2 is used on Monday and Thursday, SSR on Tuesday, TDF-TF1 on Wednesday and Sunday, TDF-FR3 on Friday and RTBF on Saturday. TV5 will not be scrambled until mid/late summer. The channel is for feeding via 12 GHz terminals to cable networks and is also available on ch. E56 (Brussels). There are rumours that BFBS propose to instal transmitters in Belgium.
UK: Mention was made of a pirate station, "Second City Vision", in the June 1983 column, operating in the Birmingham area on ch. 40 after the closedown of the local BBC-2 station. There have been press reports recently of activity over a five mile area, using the name Telstar Television. The authorities have been trying to close down the aspiring broadcaster, whose output includes current feature films etc. No reader has reported reception of the station, whose location appears to be the Edgbaston area. If it's still around, a good time to try would be at Easter - Bank holidays seem to be favourite times for this activity.
Luxembourg: The RTL-Plus channel came into operation on January 2nd. It's intended for W. Germany but is being carried on Belgian cable systems. Start of programmes is

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Tropospheric reception from Denmark ch. E8 on December 4th, showing progressive reduction of the bandwidth from 6 to 3 to 2 MHz to reduce interference, using the equipment described in the February-April 1982 issues of Television.

1727 local time ( 1657 Sunday) with closedown at $2230 /$ 2300 ( 2400 Friday/Saturday). The system B PAL transmissions are on ch. E7, the test pattern carrying the identification "KANAL 7" at the top and "RTL PLUS" at the bottom. It's expected that ch. E24 will also shortly be used.
Denmark: The experimental South Jutland service TV Syd will continue until next year. If successful, this could lead to full regional Danish TV. The main transmitter is at Sonderjylland on ch. E7 with 60 kW e.r.p.

## Satellite News

Luxembourg has applied to the ITU for permission to use four channels at $10.7,10.95,12.5$ and 12.75 GHz . The satellite would be privately funded and able to provide reception using 1 m dishes. Norway, Sweden and Finland have reached agreement on the Tele-X satellite project. Ireland is inviting organisations to tender to provide the proposed Irish satellite TV service. It's hoped that this will be on air in 1987, with five channels.
The Indian Insat 1 b satellite is apparently giving good results. There will eventually be two TV channels, with direct to community receivers as with the 1975-6 SITE experiment via ATS-6.

The North American conference on satellite services has confirmed eight slots for US DBS use in the 12 GHz band, between $61.5^{\circ}$ and $175^{\circ} \mathrm{W}$. These include spot beam coverage of Alaska and Hawaii. The conference decided on a service area signal strength of $-107 \mathrm{dBW} / \mathrm{m}^{2}$ - the USA had sought $-105 \mathrm{dBW} / \mathrm{m}^{2}$. The USA had also hoped for more slots to cater for time zone variations. STC in the USA plans to commence a five-channel 12 GHz service this autumn covering the N.E., and hopes that domestic reception via 2 ft dishes will be possible. Due to congestion, domestic 4 GHz craft are to use a reduced orbital spacing of $2^{\circ}$. As a result, receiving dish specifications will need to be tightened. Domestic dishes in the USA range from 6-12 ft and are usually of the prime focus type, with a comparatively wide beamwidth. Whereas an 8 ft dish will work well with the present $4^{\circ}$ spacing, with minimal adjacent satellite/channel interference, a $2^{\circ}$ spacing could lead to problems, particularly with dishes of less than 8 ft diameter.
The US government has approved funds for the VOATV service to Europe from the TDRS- 1 satellite at $41^{\circ} \mathrm{W}$. Transmissions are expected to start this May/June. The Home Box Office company has apparently acquired the use of an ex-US DOMSAT which is being moved to approximately $40^{\circ} \mathrm{W}$, with test transmissions expected to be in progress by the time you read this and a service starting in March. The Television Entertainment Group, a
consortium consisting of Goldcrest (UK) and four US firms (including HBO), hopes to use this satellite to provide a 24 -hour programme of films, sport, pop music etc. The HBO/VOA services would be in the 4 GHz band, intended for cable distribution.

## From our Correspondents . .

Bud Lloyd Bennett, now in Bahrain, is using two stacked long Yagi aerials - Dutch Kamco type assembled in Kuwait - and a 40 dB head amplifier to receive the 714 MHz signals from the Ekran satellite at $99^{\circ} \mathrm{E}$. He says that reception is weak, though improvements are being worked on - including a notch filter to remove Bahrain ch. 55 which causes interference.

Nick Harrold has built a 12 GHz unit, to the basic Chris Wilson/Grahame Harding design featured in the September 1982 issue, and has received OTS using an 8 ft . petal dish (who said a petal wouldn't work at 12 GHz ?). The Satellite plc programmes are scrambled, but with sync reinsertion a viewable picture can be received. The coded sound is impossible to decipher however. Some days after this reception the channel was transferred to the ECS-1 satellite. TDF-FR3 is still present via OTS, with the cogwheel type scrambling illustrated in the January 1984 issue (page 126) and noisy SECAM. For 4 GHz reception Nick is now using a commercial $110^{\circ}$ LNA. He reports that AFRTS is received with good colour from the $1^{\circ} \mathrm{W}$ Intelsat. RTM (Morocco) has moved to the $31^{\circ} \mathrm{W}$ Intelsat (from $27.5^{\circ} \mathrm{W}$ ) and is much weaker.

We've received another letter from Mel Thurlbourne in the Falklands. He reports that TV signals are now being received, generally from 1800-2100 local time, with the aerial system pointed to the north. They are on chs. A2 and 3, most likely via SpE. Charlie's Angels and Kung Fu have both been seen - in colour and dubbed in Spanish! An Argentinian news programme has also been seen. The reference oscillator in the decoder needed a tweak to obtain correct colour lock, and since then things have been going well. Stanley residents use VCRs, to the UK system I. The Cable and Wireless Intelsat link is now in operation, but for communications only.

Bill Cotterill (Tipton, W. Midlands) received most of Europe, plus Jordan ch. E3, via SpE during the summer months, using the TV-DX system described in the Feb-ruary-April 1982 issues, a Hugh Cocks upconverter/ preamplifier, and a wideband Band I aerial (Ian Beckett's design, see June 1976 issue). For u.h.f. he uses a Jaybeam JBX21 array, also a system L-I converter (see February 1983 issue). The early December tropospheric opening produced excellent signals from France, Holland and Belgium.

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# The CED Video Disc System 

## Derek Snelling

The RCA CED (capacitance electronic disc) video disc system was launched on the US market on March 22nd, 1981. It was said to have been the biggest product launch in business history, and was the outcome of substantial research and development expenditure by RCA over a period of many years - RCA are reputed to have invested more in their video disc system than in the development of the NTSC colour system (shadowmask tube and all). RCA's work on video disc systems started in the early sixties, and the decision to go ahead with the capacitance system was made in 1978.

Acceptance of video discs in the USA was slow to start with, though greater success has been achieved more recently following price reductions. The CED system was launched on the UK market last October, with discs manufactured by RCA and players produced by Hitachi.

## Basic Principles

The information on the CED disc is carried in a Vshaped spiral groove and consists of frequency-modulated capacitance variations. These variations take the form of vertical undulations along the track. Fig. 1 shows the principle. The disc is 12 in . in diameter and is housed in a protective plastic caddy. It's removed from the caddy by the player and is never touched or removed from the caddy outside the machine. The caddy contains brushes that clean the record automatically upon insertion and removal.

The disc is conductive and is tracked by a diamond stylus (for durability) to which a metal electrode is attached. Thus the disc forms one capacitive plate and the metal electrode the other, the groove undulations producing capacitance variations of the order of $10^{-4} \mathrm{pF}$. These undulations are approximately $850 \AA$ deep (it's their frequency that varies). The electrode tip is typically $2,500 \AA$ thick and the diamond tip straddles several undulations at once - it's important to appreciate that the diamond tip is merely a carrier for the electrode and doesn't "play" the disc in the way that an audio stylus does. The plastic disc is made conductive by adding carbon to the compound from which it's pressed.

The stylus tip electrode is connected to a tuned line resonant at 890 MHz . This forms part of what is called the resonator assembly, which also includes an 896 MHz oscillator whose output is coupled to the stylus driven line. The circuit is shown in Fig. 2 and the action in Fig. 3. Q101/C105/C106/L105/C108 form the 896 MHz oscillator whose output is coupled by L103/4 to L102, the stylus driven line. The capacitance variations at the stylus tip alter the resonant frequency of this line slightly, as a result of which the position of the 896 MHz oscillator's signal shifts up and down the trailing slope of the 890 MHz response curve. The result of this is amplitude modulation of the 896 MHz carrier. Since the depth of the undulations in the groove is constant, the amplitude modulation is constant. But the rate of the amplitude modulation varies, thus producing an f.m. output.

L101 couples this amplitude modulated carrier to the peak detector diodes D101/D102. C103 filters out the
carrier and the result, at the base of the buffer transistor Q102, is an f.m. output carrying the information stored in the disc's groove.

This is fed to a preamplifier before being sent for splitting into audio, chroma, luminance and DAXI information - all the signals are f.m. An automatic fine tuning signal is taken from the preamplifier circuit to varicap diode D103 to maintain the average centre frequency, i.e. to ensure that the 896 MHz signal remains at the centre of the falling slope of the basic 890 MHz response. The detected output from the arm must now be processed.

## Processing the Sound Signals

Two subcarriers, 711 kHz and 898 kHz , are used for the sound signals. A block diagram of the sound processing system is shown in Fig. 4. Filters separate the two signals, the first of which carries mono sound with a mono disc or $L+R$ (the same thing really) with a stereo disc and the second $\mathrm{L}-\mathrm{R}$ with a stereo disc. On bilingual discs or discs


Fig. 1: Cross-section showing the stylus resting on the base of the V-shaped groove. The groove undulations, which cause capacitance variations, vary in frequency.


Fig. 2: The resonant circuits and detector used to retrieve the f.m. signal from the disc.


Fig. 3: Principle of signal retrieval. The frequency of the 890 MHz resonant circuit varies with the disc capacitance variations, giving rise to an amplitude moduated carrier. The amplitude modulation varies in frequency.


Fig. 4: Block diagram of the sound processing system.
with two sound tracks the main track is at 711 kHz and the subsidiary one at 898 kHz . The bandwidth is $30-15,000 \mathrm{~Hz}$, and the signal-to-noise ratio can be improved by 20 dB by the use of CX noise reduction, which is a feature of some discs.

## Video Signal Processing

The video signal processing system is shown in block diagram form in Fig. 5. A $1 \cdot 5-9 \mathrm{MHz}$ bandpass filter passes the composite chroma/luminance signal to the video f.m. detector. The bandwidth of the demodulated luminance signal is approximately 3 MHz , with the interleaved chroma signal on a 1.52 MHz subcarrier - near the centre of the luminance bandwidth. The non-linear aperture correction system is used to eliminate phase modulation of the video signal by the 711 kHz sound carrier. The technique used is to filter out the 711 kHz component, invert it and then add it back to the video signal to provide cancellation.

The dropout compensation can replace up to three lines before signal degradation becomes noticeable.

A CCD (charge-coupled device) comb filter is used to
separate the luminance and chrominance signals - this is the standard PAL delay line technique, though in solidstate form. Because of the wide bandwidth of the CCD filter the chrominance output contains low-frequency luminance information. This is separated by a 1 MHz lowpass filter and added to the luminance signal.
The 1.52 MHz chroma signal obtained from the 3 MHz low-pass filter is mixed with the output from a 5.95 MHz voltage-controlled crystal oscillator to obtain a 4.43 MHz chroma signal which is then phase altered to the PAL standard and added to the luminance signal for sending to the u.h.f. modulator. This produces an output for feeding to the TV set's aerial input socket.

A 4.43 MHz crystal oscillator and phase detector are used to control the 5.95 MHz oscillator. The phase detector's output is also used to operate the jitter servo circuit, whose purpose is to maintain a constant stylus-to-groove velocity in order to compensate for such things as warped or eccentric discs or an off centre hole. The result of such defects would otherwise be varying colour or horizontal instability of the picture. In addition, the outputs from the 4.43 MHz and 5.95 MHz oscillators are mixed to produce a 1.52 MHz clock signal for the CCD comb filter i.c.

Another output from the 3 MHz chroma low-pass filter. is the DAXI (digital auxiliary information) signal, which is recorded on lines 20 and 333 during the field blanking period. This information is used to tell the machine if the stylus is tracking correctly, whether a mono/stereo or bilingual disc is being played and also, by checking the field number, whether the stylus has kicked back one or more grooves or, in visual search, whether the stylus is kicking forwards or backwards correctly.

## The Mechanics

The arm is a lateral tracking one driven by a d.c. motor. To ensure that the stylus remains centred in the groove, the position of the output signal coupling flylead is detected: if this moves off centre from the arm, due to the arm being ahead of or behind the stylus, an error signal to


Fig. 5: Block diagram of the video signal processing system.
correct the arm position is generated. Fig. 6 shows the arm servo system.

The disc rotates at 375 r.p.m. on a turntable driven by a three-phase direct-drive d.c. motor containing an eightpole rotor. The motor does not need to be locked to the mains frequency. Motor rotation is detected by a printed coil which generates a frequency proportional to the motor speed. This is used to control the motor speed three Hall elements generate the three-phase drive current.

The arm assembly with the cartridge and stylus is interesting. It contains the stylus position sensors and three coils - the kicker, lifter and jitter coils. The kicker coil makes the stylus jump one or more grooves either forwards or backwards. It works by pushing (electromagnetically) on a small permanent magnet mounted near the stylus. It's used during search, pause and to nudge the stylus if it sticks in the groove. The lifter coil is used to lower the stylus on to the disc - only the stylus is lowered on to the record, not the arm assembly, unlike an audio record player. The stylus is also lowered on to a cleaning pad each time a disc is removed from the player. Power is required to lower the stylus: in the event of power failure the stylus rises automatically to prevent possible damage to the stylus or disc. The jitter coil moves the stylus forwards or backwards slightly in the direction of the groove, momentarily altering the relative stylus-to-disc speed to compensate for wow and flutter.

## Practical Points

So much for the basic player. Now for a few practical points on the Hitachi range of models. There are three machines, the VIP101P, VIP201P and VIP202P. The VIP101P provides mono sound reproduction only. Features include pause (without picture) and $\times 16$ forward search. The VIP201P and VIP202P provide mono, stereo or bilingual sound reproduction (depending on the disc) with $\times 120, \times 16$ and $\times 4$ forwards and backwards search, pause (without picture) and step (a sort of freeze frame). The VIP201P has eleven-function remote control while the VIP202P has an optional wired eleven-function remote control system.

The $\times 120$ visual search is not visual search in the same sense as with a VCR. In this mode the arm travels across the disc rapidly, the stylus being lowered at intervals to sample the picture for a few fields - a sort of peep search.

The position of the arm on the disc is indicated by an LED fitted to the end of the arm. It's viewed through a window along the front of the machine. There's a rough scale in minutes along the window, to give an idea of how much playing time has gone - the maximum playing time is 75 minutes per side. Whether side one or side two of the disc is being played is also shown on the front: this is detected by a microswitch at the rear of the machine and is necessary because the disc can be inserted in the caddy either way up.

Once the machine detects that a disc has been inserted and the caddy removed - by correct sequential operation


Fig. 6: The arm servo system.


Fig. 7: Some of the mechanical aspects of the player discussed in the text.
of the door flap microswitch and a microswitch at the rear of the machine - the disc is lowered to the playing position and the arm moves across to the start of the disc. A metal plate attached to the arm passes between the two halves of a photo-interruptor: the plate has a hole in it, the position of the hole determining the set-down point. This plate can be adjusted using a test disc.

The cartridge/stylus can easily be changed by removing a small plastic cover on the top left of the machine, unclipping the stylus cover, pulling out the old assembly and dropping in the replacement. Also housed beneath this cover is a transit screw and spacer. These must be fitted whenever the player is transported, and removed prior to its use. The arm assembly complete with coils, tuned line, oscillator and preamplifier is considered by Hitachi to be a non-serviceable item that should be replaced complete.

Apart from these components the rest of the electronics are mounted on a single large panel in the top of the machine - the top can be hinged up and locked in the vertical position for servicing. A word of caution here. If the machine is operated in this position with a disc in place, you could well drop something on the disc, ruining the disc and possibly the stylus. If it's necessary to operate the machine with the top off, remove the cartridge/stylus assembly and trick the machine into loading by operating the two microswitches and the photo-interruptor in the correct sequence.

When you press reset, the arm returns to the rest position and the disc unloads ready for removal. If the disc is not removed, it will be lowered back into the play position and rotated after a few minutes. This is done to prevent the disc warping due to heat in the machine. The disc can be unloaded by pressing reset again.

## Sound and Picture Quality

What does all this technology give us by way of picture and sound? Well, the quoted bandwidth is not significantly different from that of a VCR, and having seen the picture quality I'd put it at about equal to the better VHS machines, no sharper but with perhaps slightly better noise performance. The sound is of course vastly superior, though I've not yet heard the new Panasonic machine with hi-fi, helically-recorded sound. Although the price is half that of a VCR, it doesn't of course record. Neither does it have the improvement in picture quality and the extra features available with LaserVision, though you have to pay extra for all that.

# Teletopics 

## 8 mm VIDEO

The 8 mm video system, which has been developed for use in light-weight camcorders, is to be introduced by a number of firms later this year. Several firms have released details of their plans following Kodak's worldwide announcement of its intention to enter the field with the Kodavision series 2000 video system. Kodak have also gone into the video tape market, with a comprehensive range of open-reel and cassette tapes. Kodak's tape is being produced by TDK while the camcorder and its associated equipment will be manufactured by Matsushita.

Matsushita camcorders will also be sold by General Electric (GE) in the USA, where RCA will be selling Hitachi manufactured camcorders. Sanyo have also announced their intention to enter the 8 mm camcorder market in the USA and have already exhibited prototypes. In Europe, Philips have brought forward plans to launch their VKR 85008 mm camcorder, with a possible launch date as early as June.

The Kodavision camcorders weigh about 5lb each (there are two models), are easy to carry and use, and share several features including a fast $f / 1 \cdot 2$ 6:1 power zoom lens, a $\frac{1}{3} \mathrm{in}$. Newvicon pickup tube, and automatic white balance to adjust for colour temperature variations. The two camcorders also feature fast forward and reverse, five times visual search, and an electronic viewfinder - the latter provides a miniature monochrome TV display. The review feature enables the user to replay the last four seconds of the previous recording, and both camcorders incorporate automatic exposure control. Model 2200 features manual focus, two record/playback heads and a stillframe capability. Model 2400 features an autofocus lens with manual override, three heads for jitter-free still frame and frame advance, and provision to "write" the date on the tape as it's being recorded. The 2400 also has pushbutton fade in/out control and backlight control.

A key item in the Kodavision system is the cradle, which turns the camcorder into an easy to use playback device. To play an 8 mm tape, the camcorder is inserted in the cradle which provides standard video and audio outputs. The cradle also serves the function of camcorder storage, and will charge the battery while the camcorder is in situ - the cradle can also be used to charge a separate battery to extend the equipment's recording time. In addition, the cradle is designed to include the optional


The Kodavision camcorder Model 2200.

Kodavision tuner/timer, which has 105 channel capability and offers twelve preset channel positions. The cradle is compatible with existing $\frac{1}{2} \mathrm{in}$. VCRs so that the material recorded on 8 mm tape can be transferred to $\frac{1}{2} \mathrm{in}$. tape and vice versa. Most cradle functions can be controlled by a remote control unit that comes with each cradle.

The Kodavision system is due for release in the USA this summer and in the UK this autumn. UK prices have not been decided, but on a rough conversion of the US prices announced you could expect to pay some $£ 850$ for the basic 2200 camcorder, $£ 1,000$ for the $2400, £ 120$ for the cradle and $£ 190$ for the tuner/timer.

JVC, who have also developed 8 mm equipment, maintain that the goal of compactness is easier to achieve using standard $\frac{1}{2} i n$. VHS tape. JVC's Video Movie VHS camcorder (not Victor Movie as we called it last December) is now in production and has been launched in Japan. Shipments to the USA are due to start this spring and to Europe during the summer. It uses a VHS-C E30 cassette, an ultra small drum and a new parallel loading system. The pickup tube is a $\frac{1}{2} \mathrm{in}$. Saticon.

It seems therefore that home video movie enthusiasts will soon have on offer at least two competing systems there's also Betamovie, which was first to appear. As regards price and weight, there will probably be little to choose between them.

## DBS CO-OPERATION

Last month's leader discussed some of the problems of starting a UK satellite TV service for direct domestic reception. Since then, high level talks have been held at the Department of Trade with representatives from the BBC, the IBA, the Independent Television Companies Association and United Satellites with a view to saving the project from collapse. Senior government officials from both the Home Office and the DTI were present. The BBC has proposed a compromise plan under which the two broadcasting authorities would share the costs of a four-satellite system providing three TV channels. There are both legal and practical problems, which are being discussed by a tripartite working party headed by the managing director of the BBC's DBS operation Bill Cotton, the IBA's director general John Whitney and London Weekend Television's chairman Brian Tesler.

## STEREO TV SOUND

The BBC's experimental stereo TV sound transmissions from the Crystal Palace transmitter have been mentioned before in this column. The outcome was that whilst a second f.m. sound carrier, as used in W. Germany, could provide a largely satisfactory stereo sound service with terrestrial TV a digitally modulated second sound carrier might prove to be a more attractive solution. The advantages of the digital approach have since been confirmed by tests carried out at the Wenvoe transmitter - this area was chosen because the nearby mountains can cause severe multipath propagation (ghosting) and it was considered important to establish that digital sound signals can be received satisfactorily under such conditions.

These tests have given very encouraging results. The effect of multipath reflections was found to be very small, the digital signal providing excellent stereo quality even in areas of extremely low signal strength where the picture was badly impaired by noise. It was also found that the additional signal passed satisfactorily through the fivestation relay chain used to feed one of the remote valleys. The conclusion is that the BBC's digital system is fully
viable. It uses a bit rate of about $700 \mathrm{kbit} / \mathrm{s}$ (sufficient for two high-quality sound signals), the phase modulated carrier being set at about -20 dB with a sound-vision carrier separation of about 6.55 MHz .

A further full scale trial from Crystal Palace on BBC-2 is planned to ensure that the system will not give rise to compatibility problems with the very wide range of monophonic TV sets in use. Discussions are being held with industry, the IBA and the Home Office to achieve an agreed UK standard, but it seems that regular stereo TV sound transmissions in the UK could be some four years away.

## SKY CHANNEL

Subscribers to Radio Rentals' cable TV network at Swindon are now able to receive five hours a night of programmes from Satellite Television ple broadcast via the ECS satellite. The transmissions can be received on a 10 ft dish and the cable network operators pay Satellite Television ten pence a month per subscriber. It's expected that the service, which at present consists of mainly American programmes, will shortly be available via other pay-TV cable networks. Satellite Television's main source of income comes from advertisements carried on the transmissions, which are also taken by cable networks in Norway, Finland, Switzerland and Malta. The uplink is from the PO tower in London.

## NAMES

Morphy-Richards is a well enough known brand name that's never before appeared on a TV set. There are now two Morphy-Richards 12 in . monochrome portables however, the T730 and T739. The latter in addition incorporates an LED clock with 59 -minute sleep feature. Some more brand names that might turn up on the service bench in the future. Saisho, sold by Dixons, uses Panasonic chassis; Triumph, sold by Currys, uses Toshiba chassis; Solarvox, sold by Comet, uses ITT chassis.

## TRADE RESULTS

Figures for the third quarter of 1983 show CTV deliveries ahead of 1982 by 6.5 per cent. For the first time there was a fall in VCR deliveries, of 5.2 per cent, possibly reflecting the effects of the Japanese-EEC import limitation agreement. Monochrome portable deliveries increased by 23 per cent and an even greater increase in deliveries of teletext equipped sets occurred.

During the first nine months of 1983 CTV deliveries increasd by 20 per cent, with imports increasing from 31 to 35 per cent. It appears that there was a surge in imports towards the end of the year, consisting mainly of Grundig and ITT sets from W. Germany. There's been a change in the market, with imports accounting for a greater proportion of large-screen CTV deliveries than previously.

## VCR NEWS

The VHS system has been given a significant boost with the announcement that leading US TV manufacturer Zenith Radio will in future be marketing VHS instead of Beta machines. Until now the VHS system has held 75 per cent of the US market.

Two keenly priced V2000 system VCRs have been released by Grundig. The 1600 is a basic, simple to operate top-loader with a suggested price of $£ 369$. The two-speed Model 2080 is able to provide up to 16 hours' recording/playback time in the half-speed mode. Features include instant record (single button operation), a go-to
facility, freeze frame with manual or auto advance, and an eight-programme capacity which can be set up to a year in advance. The suggested price is $£ 479$.

Heron Electronics have introduced a playback only VHS machine. It's manufactured by Funai and will be sold under the Ingersoll brand name at a suggested retail price of just under $£ 300$. Heron's managing director Ron Sulkin comments that with the large amount of prerecorded material now available there's a need for an inexpensive player-only machine.

The latest addition to the Ferguson range is the 3V38, a slim front-loader with a suggested price of $£ 429$. The specification is similar to the 3 V 35 ( $£ 489$ ) but without remote control. ITT's latest Model VR3605 also retails at $£ 429$. Panasonic's latest models are the NV370 (£516.50) and the two-speed NV688 ( $£ 633 \cdot 50$ ).

## SCOPEX BACK IN PRODUCTION

Production of the Scopex Model 14D15 double-beam oscilloscope and the 14D10 model with TV delay system for line selection is now in full swing again, only a few weeks after the acquisition of Scopex assets by Bridage Scientific Instruments and the formation of the new company Scopex Electronics Ltd. Enquiries should be sent to Scopex Electronics Ltd., 63-65 High Street, Skipton, North Yorkshire BD23 1EF (0756 69511).

## EXPANSION IN WALES

Matsushita (Panasonic) have announced a $£ 1 \mathrm{~m}$ investment plan for their Cardiff TV factory, where CTV production is being increased from 600 to 1,000 sets a day. About twenty per cent of the plant's output, which includes radio tuners marketed under the Technics brand name, is exported.

GEC-Hitachi are to start production of 14 and 16 in . colour sets at their Hirwaun plant. Two new production lines will be installed to produce the small-screen models. At present about a third of the UK CTV market of $3 \cdot 1$ million sets a year consists of the smaller screen models, a high proportion of which are imported.

## EAST CORNWALL CATALOGUE

A new mail order/trade catalogue is available from East Cornwall Components. In addition to the usual items there's a comprehensive listing of i.c.s, multisection capacitors, test equipment and tools. A retail shop is also in operation at the mail order address - 119 High Street, Wem, Shropshire SY4 5TT (0939 32689).

## HEAD CLEANING

A new head cleaning cassette for use with VHS machines has been introduced by Bib. It can be used wet or dry and incorporates a new non-abrasive spun-bonded polyester cleaning tape made to Bib's specification. The cleaning time is ten seconds and the cassette provides 35 cleanings, i.e. about four years' average use. Bib recommend regular head cleaning after 40-50 hours' playing time to remove dust and oxide particules. The new cleaner has a recommended retail price of $£ 9.98$ including VAT.

Issue 9 (December 1983) of Ferguson Feedback contains a detailed article on head cleaning. Thorn recommend cleaning by hand, using chamois leather and either isopropyl alcohol (IPA) or Isceon (MMV3601). The chamois leather should be wrapped tightly around a finger and moistened with the cleaning fluid. Then clean the heads and surrounding area by rubbing backwards and forwards across each head six or seven times (don't clean
in the vertical direction or with the heads in motion). Apply sufficient pressure for the head profile to be felt through the leather. Finally rub backwards and forwards six or seven times with dry chamois leather. Thorn comment that the main cause of poor head cleaning is the application of insufficient pressure to the head face.

## MOVE TO REAR PROJECTION TV

Mitsubishi have added a 40 in ., rear-projection set, Model VS400R, to their range of projection sets on sale in the USA. The set features a wide viewing angle ( $120^{\circ}$ ), high picture brightness ( 180 foot lamberts) and a 139 -channel frequency-synthesis tuning system. A six-element glass lens is used instead of the conventional three-element plastic one. Mitsubishi officials are convinced that there will be a significant move from front-projection TV sets to rear-projection types.

Sanyo have launched a 46 in . rear projection set, Model CVP9110T, in the UK. The set features a stereo sound system delivering up to 10 W per channel and uses three 7in. tubes each with its own lens. A suggested price of $£ 2,700$ is quoted.

## ITT RECEIVER-MONITORS

ITT have launched two new 14 in . colour receiver-monitors to meet the increasing demand from microcomputer
users. The basic RL2301/1 has RGBS inputs while the RL2301/M also has provision for a PAL composite video input and loop-through facilities. Both sets are equipped for off-air reception. ITT comment that the use of RGBS (RGB plus sync) inputs provides superior colour graphic displays since the video input does not have to be decoded.

## DIGITAL TV TRANSMISSION

The BBC's Engineering Research Department has been carrying out tests on methods of reducing the information rate required for digital television. It is generally assumed that a basic information rate of 216 million bits per second ( $216 \mathrm{Mbit} / \mathrm{s}$ ) is required for a TV signal consisting of separate brightness and colour components. The BBC has recently carried out a field trial using a system that reduces this rate to less $140 \mathrm{Mbit} / \mathrm{s}$ - the component signals were passed through the British Telecom $140 \mathrm{Mbit} / \mathrm{s}$ digital circuit between London and Birmingham. Two high quality stereo sound signals, using the BBC's NICAM-3 digital coding system, were also transmitted over the circuit. There was negligible loss of picture quality. During earlier experiments composite PAL signals were passed through a digital London to Birmingham circuit. The BBC's next target is a further bit rate reduction to $53 \mathrm{Mbit} / \mathrm{s}$, the eventual target being a bit rate less than $34 \mathrm{Mbit} / \mathrm{s}$.

## Vintage TV: The Ferguson 841T

Vivian Capel

Thorn have for many years been the UK's leading indigenous TV manufacturer. The firm did not enter the TV field at the outset however, and the earliest model appears to have been the post-war Ferguson 841T. There were two versions, with 9 in . (MW22-7) and 12in. (MW31-7) tubes - the e.h.t.s were 5 and 6 kV respectively. They were single-channel, t.r.f. sets for reception of the Alexandra Palace transmissions. The design seems to have owed something to radar practice, with lots of EF50 valves (eleven) and a transitron oscillator as the field generator.

This was one of those models that had to be treated with respect by the service engineer, the e.h.t. being mains derived. Though the voltage was low by modern standards, it packed a hefty punch and could be lethal because of the high current that could be passed. The chassis was mains isolated however, by a single transformer that supplied all the chassis' power requirements. This had a centre-tapped h.t. secondary winding (see Fig. 1) which fed an FW4-500 full-wave rectifier. There was a 4 V filament winding for this and a separate 6 V winding for


Fig. 1: The Ferguson 84it's power supply circuit.
the parallel connected valve (plus c.r.t.) heaters. Then there was the e.h.t. winding, plus a bit for the HVR2 rectifier's filament. The arrangement of the e.h.t. rectifier circuit was a bit unusual. The rectifier was connected to the low-voltage end of the winding, with its anode returned to the h.t. line. Thus the h.t. voltage was added to that obtained from the e.h.t. winding. A string of $3.3 \mathrm{M} \Omega$ resistors was connected across the e.h.t. supply, providing a constant load to improve the e.h.t. regulation and also serving to discharge the $0.1 \mu \mathrm{~F}$ e.h.t. reservoir capacitor.

One of the reasons for the comparatively large valve complement (twenty plus c.r.t.) was the use of EF50s in the r.f. sections. These all-metal envelope valves did not have as much gain as the later r.f. pentodes that took their place. No fewer than five were used in the vision r.f. stages, the first two being common to the sound channel. A further two were used in the sound only r.f. circuits. In cases of low gain it was often necessary to replace most if not all the EF50s. They tended to suffer from loss of emission, and replacing one or two would make only marginal improvement. Another problem often encountered with this type of valve was noisy and intermittent pin contact. Fortunately in the 841 T the five vision r.f. pentodes were mounted in a row along the rear edge of the chassis, and were thus easily replaced.

An EA50 was used for vision demodulation, driving another EF50 that acted as the video amplifier (see Fig. 2). Since the c.r.t. was grid modulated, the video output at the anode of V11 was positive-going. The vision interference limiter circuit operated on a time-constant basis: large positive-going spikes caused the diode across R74 to conduct, thus short-circuiting the output. The diode used was part of an EBC33 whose triode section served as the line blocking oscillator.

An unusual feature was the use of d.c. restoration in both the video feeds, to the tube and to the sync separator. D.C. restoration in the latter path was required because the sync separator had to operate with negative-going sync pulses. This complicated the design of the sync separator quite a bit - it had to be saturated during the video part of the signal, then driven to cut-off by the sync pulses, thus producing positive-going pulses at its anode.
A.C. coupling without d.c. restoration means that the signal's d.c. conditions vary with the content of the video signal - a line with a predominantly white content will have a different mean level than one that's mainly dark, due to the different proportions of the negative- and positive-going portions of the waveform. Hence the brightness level will float and the sync pulses will move up and down. Another problem is the fact that the video amplifier must be able to accommodate larger signal excursions and be able to deliver about twice the output required with a stable d.c. level. With an EF50 used as the video amplifier, there was little gain to spare.

The use of a.c. coupling to the c.r.t.'s grid avoids the problem of tube damage in the event of failure of the video output pentode, and the inclusion of a d.c. restorer maintains the correct d.c. conditions at the grid. Many devotees in the early days maintained that grid rather than cathode modulation gave superior results, but the complications, both in the tube drive and sync separator circuits, led to the general adoption of cathode drive.

The sync separator stage used an EF50 and this was followed by a further EF50 which was used to amplify, invert and integrate the field sync pulses. Yet another EF50 was employed as the field oscillator. This was


Fig. 2: The video amplifier and sync separator circuits, with d.c. restoration in both feeds.
arranged as a transitron, with feedback between the anode and control grid and also between the screen and suppressor grids. The operation has been described before in this series. The EL33 field output valve was $R C$ coupled to a transformer driving the scan coils.

The line timebase was simple indeed in those preflyback e.h.t. days. The blocking oscillator's output drove an EL38 output pentode that operated as an amplifier rather than a switch, with simple transformer coupling to the scan coils.
The audio circuit was conventional, using an EBC33 and EL33 output pentode, but the negative feedback loop incorporated a tone control. Sound was considered important in those days!
One wonders whether there's anyone left at Thorn who can recall the 841 T . It's a far cry from the TX series!

## VCR Clinic

## JVC HR7700/Ferguson 3V23

The problem was no clock display. If the clock crystals on the tuner/timer board are red ones, the first step is to change them to blue ones. Other minor capacitor changes have been suggested but are not essential. In this case however these changes did no good despite the fact that the timer microcomputer i.c. was deprived of clock oscillations. The culprit was in fact the microcomputer i.c. itself.

## S.B.

## Sharp VC8300

It's not often that we get bounce backs, to adopt a phrase from the world of TV repair. A certain Sharp machine (VC8300) with a servo fault caused some concern however. The initial report was of a tracking fault after the machine had been in operation for a couple of hours. So it was soak tested, with the scope tied to the capstan servo. A small servo adjustment seemed to cure the problem, a fact that seemed to be confirmed by making a double check from cold the following day.
Some days later it was back with us again, this time with the complaint that the fault occurred just after switching on. Again we gave it a lengthy soak test, with the scope monitoring the capstan servo ramp and sample pulses at TP711 and TP712. This proved that long-term drift was present. The sample pulses started off high up the ramp,

## Reports from Steve Beeching, T.Eng. (C.E.I.), Derek Snelling, Mike Phelan, Les Harris and Mick Dutton

then slowly moved down over a period of three-four hours. Further checks for capacitor or other component value drift proved inconclusive, so we came to the conclusion that the trouble was due to the capstan motor. A month or two later, when the customer was getting really upset, a replacement motor arrived. A further soak test confirmed that our diagnosis was correct.
S.B.

## JVC HR7200/Ferguson 3V29

I'm not'sure whether this one has been covered before, but to avid VCR fault collectors, here's a beaut! This JVC HR7200 (Ferguson 3V29) machine had a replay colour fault: the colours were in horizontal bands, so they were obviously not phase locked. I did puzzle initially why the colour-killer didn't operate, then discovered that there isn't one fitted - pin 10 of IC403 is left open-circuit. The colour phase-locked loop is in IC403, which was soon eliminated. So was the 4.435571 MHz crystal as it measured correctly in the record mode. With the scope connected to certain points the colours would lock, giving normal playback colour bars. This proved that the a.p.c. loop could be locked. The question remained as to why it wouldn't do so.

There are two phase detectors in IC403, the a.p.c. and ident detectors. Both are fed with gated burst, at pins 14
and 13 respectively. The only difference is that the input to pin 14 goes via a $90^{\circ}$ phase shift network. A scope comparison showed that the phase shift was correct, and the only discrepancy we could detect was in certain voltage readings. Pin 1 was at 5.9 V instead of 5.6 V while pin 14 was at 6.04 V instead of 6.4 V . These could be put down to tolerances. We eventually found that C462 $(15 \mathrm{pF})$, part of the $90^{\circ}$ phase shift network in the feed to pin 14, was resistive, though of the correct capacitance value - it measured a few hundred $k \Omega$ on the AVO.

Note that the circuitry around pins 13 and 14 of IC403 is high-impedance. Thus readings here will be valid only if a $\times 10$ scope probe or a high-impedance digital voltmeter is used.
S.B.

## JVC HR2650

The complaint was that when editing by either insert or audio dub the original sound was erased. Before starting on any repair work we had to make sure that the customer was doing the right thing. On this machine the camera records sound on ch. 1 or if stereo chs. 1 and 2 . The tuner records on both channels. Now when audio dubbing or insert editing it's possible to select to over record both channels or ch. 2 only. The customer complained that with the switch in the ch. 2 dub only position and then insert editing a title on to his tape all the original sound was erased, whereas the original sound on ch. 1 should have been left alone. A cross check with a stock machine confirmed that this should be so, the ch. 1 sound remaining untouched. In the customer's machine it was erased.

A tuner recording was made and a further check carried out using the audio dub facility. Ch. 2 was over recorded and ch. 1 fully erased - but a small amount of the original sound could still be heard on ch. 2! The fault? Well, the ch. 1 and ch. 2 audio erase heads had been reverse wired in production! As a result, when an audio dub was performed ch. 2 erase went to the ch. 1 erase head, thus wiping ch. 1 , and the ch. 2 record head erased and over recorded, thus explaining the residual sound that was left on ch. 2 .
S.B.

## Defeat!

There are occasions when circumstances dictate a course other than investigative repair. Such an occasion arose just before Christmas. A local dealer sent around a Sony SLF1 portable which wouldn't thread up after inserting a cassette. Three microcomputer i.c.s control this function. One in particular scans the cassette detector switch and has an output for threading drive. Whilst it scanned the switch input, it didn't activate the threading output. Sony technical agreed that a replacement for this i.c. and/or one of the other two was in order, and after replacing two of them to no avail it became obvious that without any technical information on the starting routines a great deal of time would be required to sort the fault out. The dealer was not prepared to pay for this and I believe the machine went back to Sony. If any Sony engineer has sorted it out, I'd be glad to hear from him ...
S.B.

## Toshiba V8600 with a Hangover

Being Christmas, I suppose it had to happen. Andy did say the customer suspected that someone had spilt white wine into the machine. The whole area around IC604 and IC602 was badly corroded. Most components had to be
removed for cleaning and the wire links had to be replaced. After a good clean up the machine was tried. The pinch solenoid didn't operate. This was traced to Q642 not providing IC602 with clock pulses (derived from the PG circuits). I wonder whether Andy reordered the foaming cleanser?
S.B.

## VCR Supply Lines

The fault with a Mitsubishi HS700 (the portable machine with the built-in tuner) was that it wouldn't record. A check showed that it was working correctly so far as the E-to-E mode was concerned, but attempting a recording produced no results - not even erasure of the previous sound or picture. A point here - when dealing with a no record fault, always use a tape with a previous recording on it in order to check whether the erase circuits are working or not. This can aid diagnosis.

The fact that so much was defective, i.e. no recording, no erasure and both the sound and vision affected, led me to suspect that a voltage rail was missing - in this case the REC 9 V rail. A check showed that it started off at 9 V but over a few seconds fell to $7 \cdot 8 \mathrm{~V}$. Shorting across to the permanent 9 V , rail brought the voltage up to the correct level but didn't affect the problem in the slightest. The manual is not very well set out, as a result of which we had to spend some time trying to find the record voltage rail to the erase oscillator. Eventually a line labelled DREC 9V was found. This was at 0.2 V instead of 9 V . The problem was traced to IC 2 H 1 , which processes the sound, a replacement curing the fault. A DREC 9 V line is not found in many machines - it stands for delayed record 9 V , the purpose being to allow noise-free transitions between recordings. The sequence after record is selected is as follows:
(1) The machine laces up and REC 9 V appears.
(2) The machine winds tape back for a short period.
(3) The machine goes into the combined record/ playback mode. During this time the E-to-E mode is maintained, i.e. the machine doesn't playback the tape, the idea being to use the previously recorded control pulses while synchronising the motor speed to the incoming sync signals.
(4) The DREC 9 V line appears and the machine goes into the full record mode. With the heads synchronised to the previous picture, a noise-free transition occurs.

This fault prompted me to make a list of the various voltages found in VCRs and their purposes to help with fault diagnosis. I've used 9 V as the nominal voltage, but the rails may be at 12 V - the principle remains the same.

The basic lines are usually $18 \mathrm{~V}, 15 \mathrm{~V}, 12 \mathrm{~V}$ and 9 V . These are present all the time once the machine has been plugged in and the operate switch is on. They power such things as the clock, timer, microcomputer (usually a 5 V feed derived from one of the other rails), aerial amplifier and, in some models, the infra-red remote control receiver.

The not-PB 9 V line is present all the time during operation except during playback. It powers the tuner and i.f. strip plus associated changes to provide the E-to-E mode.

The PB 9 V line is present only during playback. It switches the signal processing circuits and the servos to the playback mode.

The REC 9V line is present only during record. In
addition to switching the signal processing and servo circuits to the record mode it switches on the bias oscillator/erase circuits.

The DREC 9 V line is present on record only, after a few seconds delay. Used for noise-free transitions between recordings.
Thus failure of the not-PB 9V line would give normal playback but no sound or picture in the E-to-E mode, and record via the video and audio input sockets only. Failure of the PB 9 V line would give normal operation except for no playback. This assumes that the relevant voltage disappears, but as regular readers will know this seldom happens. What usually occurs is that the voltage drops sufficiently to affect some parts of the circuit but not others, or enough to cause intermittent operation. Worst of all is a switching transistor that leaks, causing part of the voltage to be present all the time. This can give rise to some very obscure faults - as the machine tries to record and playback at the same time for example.
D.S.

## Grundig $2 \times 4$ Super

A Grundig $2 \times 4$ Super arrived with the complaint that it chewed tapes. The machine worked all right on rewind and fast forward, though it was a bit sluggish. On record and playback however the take-up reel failed to rotate. This created a loop of tape, as a result of which the machine entered the alarm mode. We investigated without a tape in and discovered that the reel motors were deprived of power - there was only about 4 V at the emitters of the drive transistors instead of 12 V . A check back to the power supply revealed that the 2 V zener diode at the top centre was dry-jointed!

The power supply in this machine is liberally sprinkled with small safety resistors that tend to go open-circuit at the drop of a hat - R443, R453 and R485 are the favourites. With R443 ( $390 \Omega$ ) open-circuit the machine will not switch on - the clock displays 8 s and the relay doesn't energise. This resistor is in the base circuit of transistor T443 that bypasses the relay contacts to power up the microcomputer i.c. so that the latter can give a "relay on" signal. R435 and R485 (both $100 \Omega$ ) are in the 150 V and -150 V dynamic track following output stage supplies respectively - R435 will also remove the 33 V tuning supply when open-circuit. Lack of dynamic track following shows up as noise bars in fast search (like a VHS machine). Also the voltage at the brushes above the head will be $\pm 60-80 \mathrm{~V}$ instead of approximately zero volts $\pm 15 \mathrm{~V}$.
M.P.

## Akai VS2

Failure to load was the problem. The tape would half load then retract, with the "breakdown" signal showing. We found that the brakes were not being released from either the take-up or the supply reel. The brakes are operated by a pin which is engaged in a fork on the underside of the mechanism, but the pin had disappeared - when the bottom cover was removed it dropped out of the bottom PCB. We had no further trouble after glueing it back in place.
M.D.

## Sony SLC6

A customer phoned to say that his new VCR had already chewed up three tapes. We called and found a tape jammed inside the machine. After removing this we
inserted our own tape and found that the machine laced up correctly. When play was selected however the tape went slack and the slack sensor operated. Also the machine wouldn't go into fast forward, and in rewind the tape became very taught.
This suggested a problem with the take-up spool brake, so the machine was taken to the workshop for further examination. With the machine on the bench we inserted our dummy cassette so that we could see what was happening to the take-up wheel. There was plenty of takeup torque in play, and the brake was released correctly in rewind. Removing the cassette carriage enabled us to see what the problem was. There's a "cassette in" microswitch at the base of the cassette tray, and the wiring to this should be routed via plastic clips around the edge of the metalwork. This wire was not properly in place, and was getting trapped between the take-up wheel and the reel inside the cassette when the latter was in position. The problem didn't show with our dummy cassette because this doesn't contain tape reels.
M.D.

## Ferguson 3V23

The problem was intermittent failure to eject the cassette. When the button was pressed you could hear the motor running, but there was no movement. We found that the middle cog wheel on the side of the cassette housing was quite sloppy. On removing the cassette housing we noticed that the screw which holds the spindle on which this cog wheel is mounted had worked loose: tightening the screw removed the cog wheel play and cured the eject problem.
M.D.

## Sharp VC7300

The complaint was that the machine was stuck in play. Removal of the top cover revealed that the tape was loaded. When the machine was powered the tape unloaded. Play was selected, then stop: the machine remained loaded and the loading motor made a complaining noise - it was turning the wrong way, i.e. to load. I next noticed that the main solenoid did not release when the stop button was pressed. Pulling out the plunger enabled the machine to unload.
A switch which is activated by the main solenoid via a lever changes the polarity of the supply across the loading motor. Because the solenoid plunger had stuck, the switch was applying the wrong polarity supply to the loading motor. Lubricating the plunger solved the problem. This also solves the mystery about the VC8300H (VCR Clinic, November) with its thermal fuse open-circuit. Because the solenoid wouldn't pull in, the loading switch would be in the wrong position with the loading motor running in reverse.
L.H.

## Bulbs

When checking ex-rental Ferguson 3V29/30 s̄eries machines in the workshop we've noticed that the cassette lamp seems to fail rather a lot. We eventually realised that it seems to happen if the machine is switched on at the mains with the operate switch in the on position. With the operate switch kept in the off position, the bulb failure rate was cut dramatically. A similar thing happens with the clock bulbs in the Toshiba V5470B. If the machine is unplugged for the first time in a couple of years to go into the workshop, the clock bulbs almost always fail: unfortunately there's no way to prevent this.

## Letters

## THORN TX9 CHASSIS

In the January issue TV Fault Finding feature there was mention (page 145) of chopper transistor (TR62) failure in the latest version of the Thorn TX9 chassis. My experience has been that when TR62 goes short-circuit the reason is that R 165 ( $300 \mathrm{k} \Omega$ ) has gone high in value. This resistor forms part of a sawtooth generating network connected to pin 4 of the TDA4600 chopper control i.c., the sawtooth in turn affecting the drive to the chopper transistor. The result of R165 going high in value can also be intermittent blowing of TR62.
Brian Francis, Tech. (C.E.I.),
Plympton, Plymouth.

## POOR AERIAL DESIGN

It's a sad fact that the majority of u.h.f. TV aerials installed are of the cheap (some would say cheap and nasty) variety. The average customer does not seem to be prepared to pay for the good quality aerials produced by the better known manufacturers, and I doubt whether this situation will change in the foreseeable future. There's one little thing that could be changed however, at very little cost to anyone. It would enormously improve the performance of many of these cheap aerials.

For some reason the smaller aerial manufacturers tend to make their group A arrays with a reflector that's too short and too close to the dipole. The length tends to be $260-270 \mathrm{~mm}$ when it should be 345 mm : the reflectordipole spacing is usually $70-80 \mathrm{~mm}$ when it should be 100 mm . A reflector of this sort is literally worse than useless, as it acts as a director on the lower channels.

The performance of some of these aerials at the bottom end of the channel group is disastrous. On channel 21 the polar diagram looks like a starfish: the gain drops to about 4 dB and the front-to-back ratio is $1: 1$ ! For vertical polarisation some aerial riggers throw the reflector away and mount the array about 200 mm from the top of the mast, thus using the mast itself as a reflector.

This situation is becoming acute as so many Channel Four transmitters use channel 21. Surely the offending manufacturers would find it in their own interests to make a small and cheap modification that would greatly diminish the quality gap between themselves and the "big boys"?
W. Wright, Wright's Aerials,

Micklebring, Nr. Rotherham.

## TV FAULT REPORT

The following recent fault experiences may help some other readers. The first concerns an Hitachi P27FM 12in. monochrome portable - the one with the f.m. radio - the complaint being field roll when the set was first switched on. This could be corrected by adjusting the rear mounted hold control, but further drift would occur some minutes later. Eventually the end of the control's range would be reached. Well, field oscillator circuits are usually simple and easy to repair, so I gave the owner a ridiculously reasonable quote. On removing the set's case however I discovered that the entire field timebase is contained within a KC531C i.c. A spray with the freezer confirmed
that it was at fault, replacement curing the trouble. Unfortunately the repair was more expensive than originally envisaged. Ouch!

The Philips G8 chassis has been around for many years and rarely do I find anything that surprises me. This particular one came in for tube replacement plus a general service and set up. After carrying out the work, the h.t. control was set to minimum and the set was switched on. To my surprise the h.t. voltage at the two fuses on the power supply panel was 210 V . Increasing the control's setting increased the voltage, its stability indicating that the circuit was regulating - but at the wrong point. This sort of thing is usually caused by R1368 ( $470 \mathrm{k} \Omega$ ), which is in series with the h.t. preset, or the feedback resistor R1372 ( $390 \mathrm{k} \Omega$ ), but both were perfect.

The panel was removed and appeared to be an early version (the set itself was the later 550 series/BEAB/VCR compatible version), but not quite. The panel contained aspects of both versions - the panel went through several modifications. I decided to bring it up to the later standard. The h.t. preset was changed to $22 \mathrm{k} \Omega$, with $5.6 \mathrm{k} \Omega$ in parallel, and the chassis return resistor was changed to $10 \mathrm{k} \Omega$. R1384 was changed to $4.7 \mathrm{k} \Omega$ and a diode was added in parallel with the charging capacitor C1376-I used a BY207, with the anode to chassis. Some components were of the later values while others had the original values, and it seemed that the combination wouldn't do the job. Switching on with the later values fitted and the preset at minimum produced a regulated 150 V supply, with the correct 205 V when the control had been advanced to approximately mid-way.

Intermittently changing colours was the fault reported on an old GEC hybrid colour set. These receivers are now nearing the end of their useful life, but I occasionally still come across a good example. This was one such, in excellent condition throughout, with a crisp tube and polished, unmarked cabinet. The purity was appalling, but resetting this along with the convergence and grey-scale tracking produced a good picture. I switched off and on to check the operation of the degaussing circuit, and as everything appeared to be all right I left it at that. The owner had moved before the fault appeared, so I put the fault down to this.

A week later I was back. This time I switched off the blue and green guns and watched in red. This showed that the purity was continuously varying, drifting out and then back in. I've never seen a set do this before, but the circuit is simple enough, with a couple of thermistors - the usual VA8650 and a strange three-legged one with two sections. The VA8650 fell to pieces when I touched it, so I replaced it with confidence that the fault was now cured. All connections were checked, and the double thermistor had been replaced recently. Purity and convergence adjustment produced a good picture, and the red raster no longer varied.

After another week I was back again and, on removing the rear cover, saw it immediately. The double thermistor had a tiny hairline crack through the small end section. Replacing it put an end to my visits. Apparently the crack had been too small to see on previous visits, repeated thermal cycling eventually widening it.

A Philips TX 12in. monochrome portable (Model 12B711) gave croaky sound, reminiscent of crossover distortion, at low levels. At normal listening and high volume levels the output was normal and undistorted. Speaker replacement made no difference, and no transistor leakage could be measured. The TBA120AS was replaced
to make sure, but this didn't seem to have much effect. There's a modification to cure the condition, which is present with some of these portables. Change R300 from $27 \mathrm{k} \Omega$ to $18 \mathrm{k} \Omega, \mathrm{R} 311$ from $33 \Omega$ to $56 \Omega, \mathrm{R} 312$ from $2.7 \mathrm{k} \Omega$ to $3.3 \mathrm{k} \Omega$ and R 315 from $180 \mathrm{k} \Omega$ to $120 \mathrm{k} \Omega$. The modification increases the output stage quiescent current and provided a complete cure. My thanks to Philips Service for their advice.

Another common trouble with these sets is cracks in the print around the brightness/volume/contrast control potentiometers. They are usually very hard to see, but are easily found with an ohmmeter.
Stephen Leatherbarrow,
Middleton, Manchester.

## KEEP AN EYE ON THE ROOF

Some four years ago I reported in these columns on some of the unexpected things that can happen during the course of servicing. One case I mentioned concerned a customer who'd installed his own u.h.f./v.h.f. aerial on the roof of the block of flats in which he lived. His reception gradually deteriorated over a period of time, and when he eventually went up on to the roof to try to discover why he found that no less than six other aerials had been fitted to his mast, all in extremely close proximity. They'd all been installed by bona fide aerial companies, but not one of them had asked for permission to install an aerial on his property. He had them all removed a bit sharpish!

The saga of this gentleman and his aerials is by no means over however. Some three months ago he decided that some streamlining of his array was due. Since the old v.h.f. aerials were no longer in use they were to be scrapped and, at the same time, it was decided to replace the existing mast with a sturdier one (a scaffold pole to be exact). Suitable reinforcement was carried out and the single u.h.f. array was mounted at the top of the pole, the space below being left clear for the future installation of a CB aerial and a v.h.f. stereo radio aerial. He then painted the complete pole with red and white stripes (barber shop style) to deter (he said) anyone from fitting anything to it in the interim period.

Eventually he gets his other aerials and up he goes to fit them. What does he see? Two lovely u.h.f. arrays smack in the middle of his mast! He traced one to a particular flat and, after speaking to the tenants, got the name of the firm who'd installed it. A fitter from the firm came round and said he'd resite it. That was on a Friday. On the following Monday our customer went up to trace the owner of the remaining aerial. He managed this and found it had been installed by the same firm. He also found that the first aerial had been resited on a lightning conductor, complete with a large lump of baton in the $U$ of the clamp bolt to act as a spacer!

Now I'd expect this sort of thing from a cowboy outfit, but the firm concerned is a prominent member of the National Federation of Aerial Contractors... If a customer is given a quote for an aerial installation this will include all the fittings. If the aerial is then attached to someone else's mast this means that the bracket, pole, lashing wire etc. which have been paid for have not been used. In my previous communication I suggested that managers should take a greater interest in the way their firm's work is carried out. It seems that this suggestion has not been heeded.
Steve Knowles,
London N4.

## next month in



- ALL ABOUT FIELD STRENGTH

A set's performance depends on the strength of the signal it receives, but how do you go about measuring this and assessing the results? There's much cause for confusion in this subject, with field strength being quoted in various ways and meters calibrated quite differently. Harold Peters explains what it's all about and how to relate and interpret the various figures - including satellite TV field strengths.

## - THE RIGONDA VL100

Large numbers of these 6 in . monochrome portables were sold in the UK. Many repairers are reluctant to handle them because of the Russian markings and lack of information. Malcolm Burrell took a detailed look at the innards and tried out various transistor substitutions: notes to help others deal with these interesting sets.

## WIDEBAND UHF PREAMPLIFIER

At many sites a lift in the signal fed to the set will help - the problem has become more widespread with Ch. 4 now being generally available. The important thing in most cases is low noise rather than high gain. This preamplifier, presented by Roger Bunney, uses a BFR91 transistor which is intended as a low-noise wideband device.

## MORE THAN MEETS THE EYE

Full use of the senses rather than the meter can be a great help in speedy fault diagnosis. Robert Thompson tells you how to read the signs.

## - TEST REPORT

In reviewing the B and K 467 .c.r.t. tester/ reactivator Eugene Trundle takes a look at c.r.t. failure mechanisms and the effects on these of c.r.t. design changes in recent years.

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ADDRESS

# Servicing the Sony Model KV1400UB 

David Botto

We've handled quite a few of these receivers, which are reliable, well designed and give excellent picture and sound. Just the same, you can get puzzling faults.

As with many modern sets, failure in either the power supply or the line timebase results in the dead set symptom. As the first logical step in this event is to determine in which section of the receiver the fault lies, we'll start by considering the operation of the power supply. We'll use the Sony board lettering identification system to identify sections of the chassis.

## The Power Supply

The switch-mode power supply consists of a series chopper arrangement - though Sony choose to call it a choke-coupled flyback converter. The circuit is shown in Fig. 1. It's on board D, along with the line and field timebases.

The a.c. mains input is fed via the on/off switch S901, fuse F601 (3.15AT), choke T601 and the surge limiting resistor R601 ( $3.9 \Omega, 10 \mathrm{~W}$ ) to a bridge rectifier consisting of two ERC04-06SU2 double diodes. The fuse sometimes goes open-circuit for no apparent reason, a new one restoring picture and sound. Before fitting a replacement, check whether the old one is blackened. If so, then the bridge rectifier is suspect. Note that D602's two anodes are returned to the h.t. line rather than chassis:

Q605 (2SC1942) is the chopper transistor and Q604 (2SC2230A) the driver. Q602/3 (both 2SC633A) form an astable multivibrator circuit which if left to run on its own operates at about 10 kHz . When the set is running normally, the multivibrator is triggered by pulses from the line output transformer. The frequency is thus 15.625 kHz . L603 is the reservoir inductor and C621 the h.t. filter capacitor. When Q605 switches off, the efficiency diode D606 (ERC25-06) conducts to maintain the current flow.

The control action is provided by transistor Q601 (2SA733) which senses the h.t. voltage at its base - D603 provides a stable reference at its emitter. If the h.t. voltage rises, Q601's base voltage will rise and its collector current will fall. Since the collector of Q601 is connected to the junction of R611/2 in the multivibrator circuit, the effect is to shorten the on period of Q602 and thus the on time of Q605 to compensate. The reverse occurs when the h.t. voltage falls.

Excess current protection is provided by transistors Q651/2 (2SA733 and 2SC633A respectively) which are connected together to act as a thyristor. The emitter of Q652 senses the voltage developed across R651. Excessive h.t. current will increase the negative voltage developed at this point with the result that Q651/2 will switch on, removing the drive to Q604. The circuit also provides over-voltage protection. In this event the output from D652, developed across C653, will be sufficient for D653 to conduct. Q651/2 will then switch on as before.

## Dealing with a Dead Set

So there's a dead set sitting on your bench. Is the fault due to the power supply or an overload elsewhere? To
find out you'll need an 18 V d.c. supply - we always use two PP9 batteries in series, guaranteed absolutely ripple free! It's also vital - as with all Sony sets - to have a means of controlling the mains input voltage, i.e. a variac or tapped mains transformer. In addition, a $100 \mathrm{~W}, 240 \mathrm{~V}$ bulb is required.

Removing the link (see Fig. 2) right next to Q605's heatsink isolates the power supply. Connect the negative end of your 18 V supply to chassis and the positive end to the positive side of C609. Don't connect the mains supply at this stage. Connect a meter in series with the 18 V supply to measure the current drain - start on a high current range. The reading should be around 25 mA . A digital multimeter is best for all tests on the power supply.

With an oscilloscope connected via a 10:1 probe to the collector of Q602, a waveform similar to that shown in Fig. 3(a) should be seen - of about 12 V peak-to-peak. This tells you that the multivibrator is working. If the waveform is missing, disconnect the collector of Q651. If the waveform now appears, the fault is in the overload protection circuit. Check Q651/2 and capacitor C654 examine it for corrosion. Diode D651 (1S1555) can also fail.

If there's still no waveform with the collector of Q651 disconnected, it's possible that there's a fault in the error detector circuit. Check Q601, the zener diodes D607 (RD3.9E) and D603 (RD12E-B2) and capacitors C608 and C622 (also make sure they're not corroded). This leaves only the multivibrator circuit itself. We've found this to be reliable, but if necessary check $\mathrm{Q} 602 / 3$ and diode D604 (1S1555).

If the waveform at the collector of Q651 is correct, connect the scope - still via the $10: 1$ probe - to the collector of Q604 and the emitter of Q605. The waveforms shown in Fig. 3 (b) and (c) should be seen.
To avoid damage, two quick checks should be made before connecting the mains a.c. supply. First check the waveform at the collector of Q602 again, then momentarily short the anode of D653 to chassis. The waveform should disappear and won't come back until the 18 V supply has been disconnected for a minute or so. The second check is to connect the junction of R605/6 to the 18 V d.c. supply. As RV601 is adjusted the waveform will vary - in fact at one end of the control the multivibrator will stop. These checks confirm that the overload protection circuit is working and that the error detector circuit is in order.

Note that if Q605 fails, D606, Q651 and Q652 are likely to be faulty. Knowing how this power supply works is useful since similar circuits are used in other Sony sets.

Disconnect the batteries and remove the link from resistors R605/6, then connect the mains supply via the variac or tapped transformer. Monitor the waveform at the emitter of Q605 with the scope, via the $10: 1$ probe. Increase the a.c. input slowly, watching for smoke or signs of overheating. At about 40 V a.c. the scope should show a waveform of some $170-180 \mathrm{~V}$ peak to peak and the h.t. across C621 should be about 100 V . If all is well, switch off the mains input - without altering the a.c. voltage setting and connect the 100 W bulb between the h.t. line and


Fig. 1: Circuit diagram of the switch-mode power supply. The multivibrator's main time-constant network comprises R611/C610, the control transistor Q601 adjusting C610's charging time.


Fig. 2: Location of the power supply isolating link.
chassis.
Restore the a.c. input and check the waveform at the emitter of Q605 - see Fig. 3(d). As the a.c. input is increased to 150 V the h.t. line should reach $95-110 \mathrm{~V}$ depending on the setting of RV601. Slowly increase the input to the full 240 V . The h.t. voltage should remain at 105 V . If during these tests any of the waveforms don't seem to be clean, having a lot of "fuzz", replace C609 and check C606. On rare occasions C616 dries out and causes problems.

Switch off, remove the 100 W bulb and reconnect the link (Fig. 2). Switch on with the a.c. input at a low setting and gradually increase the input. At 40 V a.c. there should be a squarewave at the emitter of Q605 and at 60 V a.c. the frequency of this squarewave should increase because pulses should arrive from the line output transformer to trigger the multivibrator. If there's no change in the frequency of the waveform, find out why before increasing the a.c. input further.

Picture and sound should begin to appear when the a.c. input is about $80-90 \mathrm{~V}$. With the full 240 V a.c. mains input, an 18 V peak-to-peak line-frequency waveform


Fig. 3: Power supply check waveforms.
should be present at thermistor TH652 and at the anode of D652.

## Timebase Troubles

If the line timebase is dead, check the output transistor Q503 (2SC1875) and the efficiency diode D503 (GH1F). If necessary, check the 12 V regulator transistor Q811 ( 2 SD 471 ) and the three rectifier diodes fed by the line output transformer T801. These are D811 (V09C), D802 (GH1F) and D801 (HF1), for the 12 V , first anode and 170 V supplies respectively. If all is well, apply a very low a.c. mains input and monitor the waveform at the collector of the line driver transistor Q502 ( 2 SC 2230 A ) to see whether line drive is present. Check that it's reaching the base of Q503. Switch off as soon as line drive waveforms are seen.

Lack of line drive can be caused by failure of the line oscillator transistor Q501 (2SA677) or Q502, also the feed resistors R514 ( $4 \cdot 7 \mathrm{k} \Omega, 2 \mathrm{~W}$ ) and R513 ( $47 \mathrm{k} \Omega$, 1 W ).

In earlier Sony receivers it was almost unknown for the line output transformer to fail. T801 in the KV1400UB does sometimes fail however. The only reliable test is by substitution. Note that the e.h.t. rectifier is encapsulated within the transformer. First make sure that R811 ( $1 \cdot 2 \Omega$, $\frac{1}{4} \mathrm{~W}$ flammable) and the two diodes (D504/5, type SIB0102 ) in the centring circuit are o.k. Also examine all the small electrolytics in the line timebase for excessive leak-
age or signs of corrosion.
We've not had many problems with the field timebase. If diode D511 ( 1 S 1555 ) fails, field sync is lost. The output transistors are Q553 (2SD669A) and Q554 (2SB649A). They rarely fail. If the sides of the picture are bowed, suspect leakage or failure of the pincushion correction transistor Q581 (2SD571) and/or its emitter resistor R584 ( $82 \Omega$ ).

## The Signals Side

The audio circuit (board A) also rarely needs attention. If the sound does fail, the first thing to do is to measure the voltage at pin 6 of the intercarrier sound i.c. (IC251, type CX095C). It should be 5.4 V . If way off, check the sound mute stages on board M2 - test transistors Q182 (2SC634A) and Q181 (2SC733) for leakage, also diode D181 ( 1 S 1555 ). If the 5.4 V reading is correct, check the audio circuit. The output transistors Q252 (2SD669A) and Q253 (2SB649A) can fail. In this event check the emitter resistors R260/1 (both $33 \Omega$, $\frac{1}{2} \mathrm{~W}$ ) and the driver transistor Q251 (2SC926A).

There are five i.c.s on the signals panel (board A). In addition to IC251 these are IC281 (M5135P) for a.f.c., the i.f. chip IC201 (CX177B) and the two decoder chips IC301 (CX108) and IC302 (CX109). IC301 contains the luminance amplifier, chrominance processing and reference oscillator stages (there's also an external transistor, Q305, in the oscillator circuit) while IC302 is mainly concerned with colour demodulation and matrixing. IC201 also feeds the intercarrier sound chip of course. These i.c.s are all very reliable.
A difficult fault to trace is intermittent loss of sound and picture with a peak white raster. A number of $0.0047 \mu \mathrm{~F}$
capacitors will be seen in the vicinity of IC201-C207, C208, C211, C213 and C216. Replace the lot!

In the event of colour problems, start by connecting the $10: 1$ scope probe to the junction of $\mathrm{C} 310(100 \mathrm{pF})$ and R313 ( $1 \mathrm{k} \Omega$ ). With a colour bar input, chroma bars should be seen. It's handy to know how to disable the colour killer: this is done by applying 4 V (use one of your PP9 batteries and a $5 \mathrm{k} \Omega$ preset) to pin 24 of IC302. If there is still no colour, go straight to the oscillator transistor Q305 (2SC403C) - there should be a 4.43 MHz sinewave at its collector. If there isn't, check the transistor and the components in this area.

The colour-difference and luminance outputs from board A can be easily checked with the scope at plug A5. The luminance signal should be present at pin 1, with $\mathrm{R}-\mathrm{Y}, \mathrm{G}-\mathrm{Y}$ and $\mathrm{B}-\mathrm{Y}$ at pins 3, 4 and 5 respectively.

The RGB output transistors Q701-3 (2SC2278) are on c.r.t. base panel C. Predominance or absence of one colour can be caused by leakage or failure of one of these transistors.

If the set is stuck on one channel, first clean the touch contacts then suspect the two i.c.s (IC151/2, both type M51231P) on touch sensor panel M1. This doesn't happen often.

We've not had problems with panel S (18V regulator, a.g.c. amplifier plus i.f. preamplifiers). If the tuner should fail it's best to obtain a new one from Sony rather than trying to repair the old one. If the tuning drifts off after the set has been on for some time, change D172 ( $\mu \mathrm{PC} 574 \mathrm{~J}$ ), the 33 V regulator on panel M 2 .

It's best to obtain all replacement parts from Sony. After carrying out any soldering on the printed boards, spray on a very thin coat of circuit varnish in order to prevent future corrosion.

## TV Fault Finding

## Reports from Mick Dutton, P. Hardy, M.S. Barakat and John Coombes

## Grundig 3402

The customer complained that the set would sometimes not switch on from cold. He mentioned that a faint hum could be heard from the back. This model is fitted with the CUC220 chassis, with full remote control. We made the set go into the fault condition by turning it on and off over a period of hours. The customer was correct in saying that a humming noise could be heard from the back - it was coming from the chopper transformer TR651. Voltage measurements showed that all the outputs derived from this transformer were present and correct, but we noticed that the standby relay. hadn't operated. Pin 13 on the tuning panel was shorted to chassis to prove that the fault was not in the remote control circuitry. This still produced no results, so we removed the relay's cover. The problem was now clear - the relay contacts were jammed open. Careful cleaning provided a cure, but to be on the safe side we decided to fit a new relay.
M.D.

## Thorn TX9 Chassis

The problem was that the picture would intermittently flicker. This seemed to be due to a fault in the tuner unit, but was difficult to prove since it was so intermittent. A new tuner unit was fitted and the set given a long soak test. After several days the fault reappeared, but once again it was hard to decide upon a specific cause. Luckily
we had another set with the same chassis in the workshop, so we tried swapping over the i.f. modules. This cured the fault on the first set, and after a time it put in an appearance on the second set. Careful examination revealed that one leg of the a.g.c. reservoir capacitor C35 $(1 \mu \mathrm{~F})$ was loose. A replacement cleared the fault. M.D.

## Philips K30 Chassis

The complaint with this K30 series set was that the background colour would change intermittently. On soak test we found that the red level was varying: with the blue and green guns switched off, it was possible to watch the red level gradually drifting up and down. With the use of freezer and a hairdryer the problem was traced to D4255 (BAV21) on the RGB output panel. The diode was going leaky when warm - it's in the red channel d.c. stabilising circuit.
M.D.

## Grundig CUC720 Chassis

Picture jitter was the complaint with this set: a few minutes after switching on, the picture would become very unstable both vertically and horizontally. We suspected the switch-mode power supply, but the output voltages remained stable during the fault condition. We nevertheless tried reducing the output voltage (R647), which made
the jitter less prominent. We also noted that the control didn't seem to have much range. When we removed the plastic cover from the print side of the power supply the reason for the fault was obvious. The BU208A chopper transistor's base connection had a white deposit around it, causing a dry-joint. Someone had been very generous with the heatsink compound during manufacture, and this had pushed right through the base pin connection. We had to remove the transistor and file its pins clean before a satisfactory connection could be made.
M.D.

## Rank T22 Chassis

The initial problem with this set was no results. This was soon traced to the $1 \Omega$ feed resistor in the BU208 line output transistor's base circuit being open-circuit. A few days later the customer complained that the set had once again gone dead. This time the $1.8 \mathrm{k} \Omega$ feed resistor 5 R 3 in the line driver transistor's collector circuit had sprung open. We thought we might have disturbed the soldering on this resistor during the previous repair, and after resoldering it the set seemed to work all right. A few days passed, then the same thing happened again.
This time we resoldered the resistor and left the set running. After scveral hours the picture went unstable horizontally, giving an effect similar to that produced by tripler breakdown. The problem was traced to pin two of plug 4 Z 2 having a bad connection. This pin makes the chassis connection between the line output and the scan drive panels.
M.D.

## ITT 1611

This set, fitted with the ITT80 $\left(90^{\circ}\right)$ chassis, would intermittently go dead or fail to switch on from cold. We found that in the fault condition the mains supply was not present at the mains panel. After removing the remote control receiver assembly we found that the mains input was present at the relay-operated on/off switch but one side of this was open-circuit. Stripping the unit down and cleaning the contacts provided a complete cure.
M.D.

## Tatung 120 Series Chassis

Loss of one colour is quite a common fault on these sets. Swapping over the RGB leads at the tube base will eliminate the decoder of suspicion. The RGB output stage transistors on the tube base panel may then need to be checked as appropriate, but the cause of the trouble is more often the feedback resistors R226 (red channel), R244 (green) and R251 (blue). They are $100 \mathrm{k} \Omega$ metal film resistors with a $2 \%$ tolerance rating, but tend to increase in value to something like $2 \mathrm{M} \Omega$.

In the event of field collapse, check the supply to the TDA1170 field timebase chip - there should be 23 V at pin 2. If this is missing, check rectifier diode D404 (BA159) and the surge limiting resistor R434 (10 ) in the line output stage. One or other is likely to be open-circuit. If there's a short-circuit in the i.c., R434 will burn and go open-circuit. This resistor is a metal film safety component (fusible) and must be replaced with the same type. J.C.

## Thorn 1690 Chassis

The complaint with this set was a common enough one - a severe hum bar which also upset the field sync. The l.t. rail
was slightly low at 10.5 V , and adjustment to 11.3 V only made matters worse. The mains rectifier diodes W8/9 were changed, then the reservoir capacitor C 70 and the series regulator transistor VT10, but the fault persisted. The error amplifier transistor VT13 appeared to be sensitive to freezer spray so a replacement was fitted. Still no difference. The zener diode W5 was then changed, apparently clearing the fault. The l.t. rail was low at 10 V however, and when reset to 11 V the hum bar reappeared.

A scope was then used to check the ripple at the emitter of the series regulator transistor. It was higher than expected at about 4 V peak-to-peak, and of 20 msec duration instead of 10 msec . This revealed that the rectifier circuit was operating in the half-wave mode, and a quick check at the anodes of the two rectifier diodes confirmed that one of them was receiving no current from the transformer. A meter then showed that one half of the secondary winding was open-circuit - it had been poorly soldered at a tag. A clean up and some fresh solder restored normal operation.
P.H.

## Bush TV350

This set came in with a faulty picture - going negative. On checking we found that the fault was in the a.g.c. circuit. The set worked o.k. with a weak signal, but with a strong signal overloading was seen. Checks on the transistors, diodes and capacitors in the a.g.c. circuit showed that they were all in order, but there was no gating pulse from the line output transformer. A resistance check showed that the pulse winding was open-circuit. Apparently the transformer is no longer available - from Mastercare anyway - so we decided to add a few turns to the existing transformer. About seven-eight turns of " 3000 " line oscillator coil wire were used, connected to terminals 2 and 3 of the transformer. As a safety measure, insulating tape was used before and after the winding. When the set was switched on it was found to be working normally. Note that if the winding is in antiphase the field locking is difficult - in this case reverse the connections of the new winding to pins 2 and 3.
M.S.B.

## ITT CVC20 Chassis

In the event of excessive width, check whether C72 $(4 \cdot 7 \mu \mathrm{~F})$ is short-circuit then if necessary check the EW modulator diodes D23 (MR854) and D24 (BYX71-350) by substitution.
J.C.

## Amstrad CTV2000

This set uses a chopper power supply with the control circuit and the chopper transistor in a single chip, IC502 (type STR451). In the event of no sound or picture, check for 103 V output at pin 2 . If this voltage is present, check the voltage at the collector of the line output transistor Q705 (2SD904). If correct at 102 V , check back to the collector of the driver transistor Q704 (2SC1756) where a reading of about 75 V should be obtained. If this voltage is absent the feed resistor $\mathrm{R} 726(1 \mathrm{k} \Omega, 1 \mathrm{~W})$ is probably opencircuit.
J.C.

## Thorn 9000 Chassis

In the event of tripping with a loud hum, check for dryjoints or open-circuit print at the mains rectifier's reservoir capacitor C702 $(400 \mu \mathrm{~F})$.
J.C.

## VCR Servicing

Part 26
Mike Phelan

The 3V24 uses dual-loop servos, like the 3V23, but in this case the phase control loops are both contained in an HA11711 i.c., as in the 3V16 mechanically controlled machine which we discussed earlier in this series. The drum motor is a direct drive, Hall effect type, also as in the 3V23.

## Servo System on Record

Fig. 117 shows the basic drum and capstan servo arrangements in the record mode. Briefly, the drum servo is locked to the off-air field sync pulses after division by two. This is done by comparing the phase of the 25 Hz pulses with that of a trapezium waveform derived from pulses obtained from a pick-up head associated with the drum motor. The divided-by-two field sync output also provides the feed to the control head. The $1,500 \mathrm{~Hz}$ FG (frequency gear) signal produced by the drum motor assembly is converted to a control voltage for the speed control loop. The capstan servo is controlled by a 32.768 k Hz crystal oscillator whose output is divided down to 21 Hz and then converted to a trapezium. This is compared to the 126 Hz FG signal after division by six. The FG signal also operates the capstan speed control loop.

## Servo System on Playback

On playback (see Fig. 118) the crystal oscillator provides a reference signal for the phase-control loops in both servos. It's divided down to 25 Hz (not 21 Hz this time). In the drum servo it's used to gate the sample trapezoid obtained from the feedback pulses provided by the pickup head. In the capstan servo the divided-down output from the oscillator is converted to a trapezoid which is gated by the pulse output from the control head. There are also shuttle search, still frame and (by remote control only) slow-motion facilities. The speed control loops operate in the same way in both the playback and the record modes.

## The Drum Servo

Fig. 119 shows the drum servo in greater detail. We'll look at the drum motor drive amplifier later. Suffice it to say for now that it has a forward and a reverse input (the latter is used for braking - the head never goes backwards!), also a control input, switched by transistor X6, so that the motor is stopped in fast forward/rewind/stop. As in the 3 V 23 , the forward and reverse inputs are driven by two operational amplifiers so that the phase and speed control voltages can be added. When X6 is on, the two control lines are grounded via the isolating diodes D8 and D9 and the motor stops. Transistor X5 is included to ensure that only one control line goes high at one time otherwise the motor drive amplifier would be damaged. A similar arrangement was used in the 3 V 23 .
The speed control loop is mostly contained in IC3 (type VC1029). The $1,500 \mathrm{~Hz}$ FG input is amplified to approximately 1 V peak-to-peak. It emerges at pin 3 and is fed
back in at pin 4. It's then squared and converted to a sawtooth whose slope is constant, determined by the setting of the drum free-running speed control R13. If the drum speed decreases, the average ramp voltage falls. This is integrated within the i.c. to provide a d.c. voltage, and is also inverted to produce a rising output to speed up the motor. Note that with this type of system if a decrease in speed is called for the forward output to the motor drive amplifier goes low and the reverse output goes high, braking the motor until the correct speed is reached.
The phase control loop is the same as that in the 3V16, so we'll not go into detail. The network connected to pin 10 of CC 1 should by now be familiar. R1 is the customer's tracking control which gives $\pm 10 \mathrm{msec}$ phase shift (equivalent to a quarter revolution of the head drum) each side of the mean point set by R2 and the linear position of the control head. The latter is adjusted with an alignment tape while R2 is adjusted on the machine's own recording. D1 conducts on record, shorting out R1 and R2. This reduces the monostable multivibrator's time-constant, which is still adjustable by R4 to set the record head switching point. IC1 also contains the capstan servo, which is why the oscillator's output is divided down to 21 Hz on record.


Fig. 117: Block diagram of the record servo system.


Fig. 118: Block diagram of the playback servo system.


Fig. 119: Main features of the drum servo.

Pin 8 of IC2 is fed with a 4.6 V reference voltage, this being approximately the correct value for the drum to free run at the correct speed. C24 and C25 form a timeconstant to maintain this voltage against short-term variations. Pin 6 of IC2 is low in record and playback (switch open) while pin 12 is high (switch closed), thus enabling the phase control loop. If the sync pulses are lost during record, sync detector X4 drives pin 6 high to maintain the drum speed and prevent hunting. Pin 12 goes low in shuttle search (search fast forward and search rewind), opening the phase control loop: either pin 5 or pin 13 goes high, connecting the appropriate presets R26/R29. These are set to give the correct drum speeds ( 1,560 or 1,425 r.p.m.) to maintain the correct number of lines per field in the shuttle search modes. In stop, fast forward and rewind pin 6 goes high so that C24/5 are charged to the reference voltage to give the minimum lock-up time when record or playback is selected.
As in the 3V23, a "drum vibration" signal is fed to pin 5 of IC4 in slow motion to correct the sideways motion of the picture. This is necessary because the different writing speeds when the tape is alternately moving and still would cause a variation which has to be corrected. For this purpose a waveform obtained from the flip-flop in the slow-still i.c. is added to the loop error voltage.

## Drum Motor Drive

Fig. 120 shows the drum motor drive amplifier circuit. As mentioned in Part 24, the resistors on this panel are of the chip type to conserve space. The motor has two sets of stator coils and a permanently magnetized rotor. Each coil is driven by a six-transistor bridge circuit. There are two Hall effect sensors (HG1 and HG2) which are switched on



Fig. 120: The drum motor drive amplifier circuit.


Fig. 121: Basic construction of the drum motor.
by the rotor magnets, also two control amplifiers (X1 and X 2 ) and a balance ciruit (X27). Only one bridge circuit is shown in Fig. 120 - the bridge circuit for stator coil 2 is identical.

As the motor rotates, HG1 produces two antiphase sinewave outputs that switch X6 and X8 on alternately. When X 8 is on, $\mathrm{X} 14, \mathrm{X} 19$ and X 22 are on: when X 6 is on X15, X18 and X23 are on. Thus the current through stator coil 1 is reversed smoothly as the motor rotates.

To increase the motor speed, X2's base current is increased. Its collector voltage falls, thus increasing the conduction of X6 and X8 (and the corresponding transistors in the stator coil 2 circuit). The output transistors X14/15/18/19 in turn conduct more heavily and the motor speeds up. To provide braking, X1 turns on and X2 off. X5 and X7 then turn on. These are connected in the same way as X6 and X8 but with reversed base connections to the Hall effect i.c. As a result, the rotating stator field reverses, and so would the motor if the process continued, but the servo action prevents this - "reverse" is just used to slow the motor.

The $0.68 \Omega$ resistor R 34 provides a convenient test point for scoping the motor current and also a slight amount of negative feedback as the earth return current for both
bridges passes through it.
This leaves the dual transistor X27 with its common emitter connection - it's a long-tailed pair. The Hall effect i.c.s are fed with approximately 5 V via R23 and R24, the output sinewaves sitting on about 4 V . Because the rotor pole fluxes are not necessarily equal, the two Hall effect i.c.s can give markedly different outputs. This would produce a rhythmic speeding up and slowing down of the head during each rotation - and bent verticals on the picture. To prevent this, the outputs from each Hall i.c. are fed to the bases of X27. As the outputs are in antiphase they cancel, leaving an average d.c. level which is what we need to monitor. If the d.c. tends to rise, the relevant half of X27 will turn on harder, reducing the supply to the i.c. Because of the common emitter resistor R31, an increase in the current flowing through one half of X27 will reduce the current flowing through the other half. This reduces the supply to the other i.c. and increases the output level from it. This balancing system ensures that the outputs remain the same. In practice the correction voltages are in the form of 25 Hz sinewaves, most of the errors coming from magnetic imbalance of the rotors.

Before we leave the drum servo, a few words on the construction of the motor (see Fig. 121). The lower drum assembly, including the motor centre, is a finely machined alloy casting containing the two ballraces for the spindle. Under this is the printed FG coil on a printed board which also carries the Hall effect i.c.s, disposed at $90^{\circ}$. The plastic encapsulated stator coils are mounted vertically on this. The assembly is covered by the cylindrical rotor, which is a steel pressing and also acts as a flywheel. The six-pole permanent magnet, of cyclindrical form, is inside the rotor, with the 60 -pole annular FG magnet on its upper end. The pick-up magnets (one N pole, one S pole) are mounted externally on the rotor, the pick-up head being mounted on the chassis. A word of warning - the ballraces are preloaded on assembly: don't dismantle the unit, as this preload is difficult to achieve on reassembly.

Next month we'll look at the capstan and reel servos.

## The Card Game is Over

Les Lawry-Johns

Some time back I mentioned the card game we played at the Call Girl at lunchtime on Sunday. Honey Bunch's partner was Sean (John). For a long time Sean suffered from a bad heart and a damaged leg. Recently his leg got much worse and he was taken to hospital. He was found to have lung trouble and didn't survive long after an operation. H.B. took it upon herself to arrange the burial and all the other things that have to be done when there are no relatives to handle them. It's now over and done with, but we are left puzzled by the vacuum that Sean left behind him. No papers, no letters, nothing. He didn't talk much, and when he did he talked so softly that few heard all he said. We know that he had been resident in the Waterford or Wexford area and that he had served in the police force there for some ten years. We also know that the magazine has a number of readers in that area. He spoke of his father, brother and dog. Perhaps someone there knows a little more about John Joseph O'Leary? If so, we should like to hear from them. He was well known and liked here.

## Testing Ultrasonic Handsets

Someone brought in a remote control handset that wasn't working. I checked it over, resoldered several suspect joints and fitted a new battery. I then realised that I didn't have a suitable set to check it on. All our new sets have infra-red remote control. The cat (Spock) was asleep on top of one of these sets. I pointed the unit at her and pushed the button. Her ears flicked. I waited a while then tried again, with the same result. This world shaking scientific test was carried out several times. We wrote: handset repaired and subjected to repeated tests on suitable receiver.

## The Philips G11

The G11 can be a bit of a pain at times. One pained me the other day. A white line across the screen testified that all was not well with the field timebase circuit or the supply to it. Normally the TDA2600 field timebase chip goes short-circuit internally and blows the 800 mA fuse on the line output panel. So, finding the fuse blown, I removed the heatsink on the TDA2600 and fitted a new chip. I then checked for shorts and fitted a new fuse. Switch on and pop goes the fuse.

I checked again for shorts. None. So I removed the chip, replaced the fuse and tried again without fitting a TDA2600. The fuse held. Fit another TDA2600. Pop. Conclusion: the i.c. was in order, the short occurring only when it came into operation. I looked at the circuit diagram and tried this, that and the other. It took this idiot some time to find that one of the two parallel-connected $1,000 \mu \mathrm{~F}$ output coupling electrolytics was dead short. I should have found it in the first place.

## Miss Spray

Miss Spray came in to tell me that her Pye 725 was playing up - the colours were constantly changing. We immediately diagnosed a faulty RGB output stage thick-
film resistor unit, and this proved to be correct. However... She had these two little dogs with her and they immediately caught the smell of Ben. They then tried to cover every vestige of such smell as best they could. After telling me her tale of woe she noticed what was going on. "You naughty boys" she snapped, "sorry Lorry".

I smiled weakly. Thank heavens they were small dogs. It took me about an hour to remove all traces of their visit.

## Mrs Plunky's G8

Mrs Plunky phoned to say that her Philips TV (G8) had suddenly lost height. As she was on her own she couldn't bring it in. Not wishing to be away too long, I grabbed a G8 timebase panel and the rest of the boxes and sped to her house. I took in only the toolbox and the panel. She showed me the picture, and although the height was indeed lacking there was also a nasty curve inwards at the right-hand side. I decided against the panel and nipped out for the spares box. Removing the rear cover, I held a mirror to the front of the h.t. reservoir capacitor: there was severe deterioration, so out it came. Unfortunately I'd forgotten to put a $600 \mu \mathrm{~F}, 300 \mathrm{~V}$ electrolytic in the box. I'd several of the $470 \mu \mathrm{~F}$ type for the G11 and one $200+300 \mu \mathrm{~F}$ 350 V electrolytic can (Pye hybrid type). The latter was too long to fit in the original position, but it stood up nicely and the clip could be fitted to keep it there. The two positive tags were moved together, soldered to the red lead, then black to the negative tag and all was well. A nice picture with full height and width.
"Who's a clever boy then?" I squawked. "Who's the best boy in the world?" Unknown to me however Mrs Plunky had returned and was standing behind me. She was giving me an odd look.
"Do you always sound like a parrot?" she queried.
"Er, well. It't not so much a matter of parrots. My wife is trying to teach this young cockatiel to speak and it's sort of catching."
"She seems to be teaching it to be rather conceited" sniffed Mrs Plunky. "Do I owe you anything for this quick little job?"

Oh dear. No one seems to appreciate me any longer.

## Round the Room Four Times

We get our share of strange tales. This young couple struggled in with their Ferguson 9600. The young man started the tale, which was eagerly taken up by the girl.
"The set goes all right for some time and then the picture goes funny" said he. "And we have to unplug it, wheel it round the room four times, then it's all right for the rest of the evening" said she. I looked at the set for some time, then asked the key question. "Clockwise or anticlockwise?"

She was struck dumb for once. "Clockwise" said he after a pause.

I turned the set up and, with the rear cover off, looked for a dry-joint under the centre section. "It curves in at the sides" he said helpfully. So I concentrated on the EW correction circuit and found one of the modulator diodes loose in its solder at one end. A quick dab of the iron with
the help of some fresh solder completed the job. When the set was turned the right way up the picture was slightly impure at one side - well, would you like being stood on end?
"Now listen" I said, with as straight a face as I could manage. "When you get back, wheel the set around the room four times anticlockwise. To unwind it, see?"

The girl nodded. The young man got the message but went along with the leg-pull. "Magnetism of the earth" he said.
"Exactly, and good luck to you both."
"When are you going to repair the set?" asked the girl.
"Already done dear. It had a cold and needed warming up..."

## DX Signal Detector/Alarm

The circuit described in this article is capable of detecting very weak TV signals and providing an audible indication that a signal is present. It was developed primarily for use with rapidly changing sporadic E propagation. The basic idea is shown in Fig. 1. The video signal itself is used to provide the sound, giving an immediate indication of signal-to-noise ratio and interference. The filter section detects the $15,625 \mathrm{~Hz}$ line frequency component of the signal, producing a switching voltage to control the video feed to the audio amplifier. In use, the channel being monitored must initially be clear of 625 -line signals - the presence of some 405 -line information will not upset the circuit's operation. Two phase-locked loops are arranged as a narrow-band filter to generate the switching voltage. System M, 525 -line signals (line frequency $15,750 \mathrm{~Hz}$ ) are also detected.

## Circuit Description

Fig. 2 shows the circuit. For the phase-locked loops, two TDA2591 (alternatives TDA2590 or TDA2593) i.c.s are used. These are fairly complex i.c.s intended for use as the sync separator and line generator sections of a TV receiver. Not all the internal circuitry is used. The sections that are used are shown in block diagram form in Fig. 3. These are the sync separator, oscillator and phase detector, i.e. the phase-locked loop, the coincidence detector whose output varies the gating of one input to the phase detector, and the pulse generator and output stages. Use is also made of the fact that the voltage at pin 4 can be employed to switch off the output at pin 3, while the burst gating/blanking pulse output at pin 7 is used for setting up.

It might at first sight appear that the output from the coincidence detector, at pin 11, could be used to indicate the presence of a signal without any further complication. The output here is similar whether the input consists of noise or a strong locked signal however. Instead, the two TDA2591s are run with slight frequency offsets: when a signal is present, an output is obtained once both circuits have locked in. The principle is shown in Fig. 4.

A negative-going video input should be used, though reduced performance will still be obtained with a positivegoing input. The video input is first filtered by R1 and C1 to reduce the noise bandwidth. As there's no need to worry about picture cogging, more filtering than usual is


Fig. 1: Principle of the DX signal detector/alarm.

## G.R. Exeter

employed. This filtering also means that there's no need to make use of the noise-cancelling circuits within the i.c.s. The video signal is then buffered by Tr1 and fed via C2 and C8 to the sync separators in the i.c.s and via R23 and C 16 to a convenient input for the TBA120S i.c.

The external oscillator capacitors are connected to pin 14 of the two TDA2591 i.c.s while pin 15 is used to set the frequency. High quality components should be used here in the interests of long-term frequency stability. The phase detector output at pin 13 is filtered and fed back to pin 15.

The output obtained at pin 3 of IC1 is applied to pin 4 of IC2 so that the latter produces an output only when pin

## Components List

| Resistors: | Capacitors |  |
| :---: | :---: | :---: |
| R1 1k | C1 1 n5 | ceramic |
| R2 4k7 | C2 0.47 | polyester |
| R3 1M8 | C3 0.1 | polyester |
| R4 1k2 | C4 4.7 | 63 V axial electro. |
| R5 82k | C5 10n | polyester |
| R6 12k | C6 4n7 | polystyrene |
| R7 100k | C7 47 | 25 V axial electro. |
| R8 2k7 | $\begin{array}{ll}\text { C8 } & 0.47\end{array}$ | polyester |
| R9 2k7 | C9 0.1 | polyester |
| R10 108 | C10 4.7 | 63 V axial electro. |
| R11 1M8 | C11 10n | polyester |
| R12 1k2 | C12 4n7 | polystyrene |
| R13 82k | C13 47 | 25 V axial electro. |
| R14 12k | C14 47 | 10 V axial electro. |
| R15 100k | C15 47 | 25 V axial electro. |
| R16 108 | C16 0.1 | polyester |
| R17 1k | C17 1 | 63 V axial electro. |
| R18 10k | C18 0.1 | polyester |
| R19 2k2 | C19 in | ceramic |
| R20 100 | C20 100 | 25 V axial electro. |
| R21 1k2 | C21 0.1 | polyester |
| R22 $820 \Omega$ | C22 47 | 25 V axial electro. |
| R23 100k |  |  |
| R24 10k |  |  |
| R25 100k |  |  |
| R26 10』 |  |  |
| R27 10, |  |  |
| All $\frac{1}{4}$ W, 5\% |  |  |
| Miscellaneous: |  |  |
| D1 | 1N4148 |  |
| Tr1, 2 | BC252B or equivalent |  |
| IC1, 2 | TDA2591 |  |
| IC3 | TBA120S |  |
| IC4 | LM380 |  |
| RV1, 2 | 47k sub. min. horizontal preset 10 k 15 mm min. PCB mounting |  |
| RV3 |  |  |
| Small speaker |  |  |



Fig. 2: Circuit diagram.


Fig. 3: Block diagram of the sections of the i.c.s used.


Fig. 4: Principle of the PLL narrow-band filter.

3 of each i.c. is high (with pin 4 at half the supply voltage, there's no output at pin 3). This output is integrated and buffered by $\operatorname{Tr} 2$, producing a control voltage which is typically less than 2 V with no signal and $3-5 \mathrm{~V}$ when a signal is present. This is an on/off output, though noise and phase jitter make the control voltage appear to
change more linearly.
The TBA120S is used to provide the switching action, the output from $\operatorname{Tr} 2$ being fed to pin 5 - in normal use, the d.c. volume control is connected to this pin. The i.c. smooths out the on/off transitions and avoids a jarring crash when a signal is detected. Another advantage of the TBA120S is the 70 dB attenuation possible, ensuring that there is no stray low-level output from the speaker.

The switched video output appears at pin 8 of the TBA120S and is then fed via a volume control potentiometer to pin 2 of a simple audio amplifier i.c. This in turn drives the speaker.

## Construction

The prototype was constructed on a PCB (see Figs. 5 and 6) and has given satisfactory operation for many months. The following points should be noted however.


Fig. 5: Component layout.


Fig. 6: Track pattern, scale 1:1.


Fig. 7. Adjustment displays.

First, in all systems using two oscillators at the same frequency there's a tendency for them to lock together if the earthing and supply decoupling are inadequate. A small amount of coupling was present with the prototype when the input was grounded, but this was sufficient for only a few Hz of common lock-in range and didn't upset the setting-up procedure.

Secondly the input is very sensitive to 15 kHz signals. If
it's not connected directly to a low-impedance source of video in the i.f. strip, it should be screened from any nearby timebase radiation. Timebase currents should be kept separate from the receiving system.

Thirdly the TBA120S's volume control characteristic has a fairly wide spread. The values of the pot-down resistors linked to pin 5 may require optimisation therefore. The device used should have internal bias resistors connected to pins 7 and 9 . All devices examined, other than ITT ones, appeared to have these: the ITT i.c.s rely on the presence of a quadrature coil to provide d.c. bias at pin 9, a difference that doesn't show when the i.c. is used as an f.m. detector.
A miniature $8 \Omega$ speaker was used, with R27 included to restrict the maximum power. Any small speaker can be used. A 12 V regulated supply capable of supplying 140 mA is required. If not already available in the set, one can be built using any of the popular i.c. regulator circuits.

## Alignment, Testing and Use

To set up the system, ground the input and loosely couple pin 7 of IC1 (11V peak-to-peak burst gating/ blanking pulse output) via suitable attenuating resistors to the video channel of a system locked to a standard broadcast signal. Adjust RV1 for an offset of approxi-
mately 50 Hz (see Fig. 7). This, as shown, can be judged by the slope of the pulse output displayed on the screen. Repeat this step for IC2, setting the offset in the opposite direction. If available, a frequency counter can be used, adjusting for $15,625 \pm 50 \mathrm{~Hz}$. If the frequency difference is too large, IC1 and/or IC2 may not be able to pull in: if too small there may be false locking problems. The prototype worked well with 50 Hz and 100 Hz offsets. Pin 7 is used as the setting-up monitoring point rather than pin 3 as there will always be an output at pin 7 of IC2.

The unit should now be working. If there are problems, a further test can be made at the emitter of Tr 2 as video signals are applied and removed. The voltage here should vary from below 2 V to greater than 3 V . The audio control stages can be tested by varying the voltage at the base of Tr 2 between 0 V and 4 V (volume control midway) with an input signal present. The audio output should be present at 4 V .
The total supply current with no audio output should be
about 100 mA , and without IC 1 and IC 2 about 35 mA .
When two strong signals are received simultaneously, it's possible that IC1 may lock to one and IC2 to the other, giving a lock-out condition. Usually however a warning of improving conditions will already have been given, so the condition should not be troublesome.

As it stands the unit is capable of monitoring one channel continuously. To cover more channels, either more receiver systems and detectors are required or alternatively the tuner could be stepped through several preset channels at a rate slow enough for the detector to respond, say three-four seconds per channel. A method of doing this was described in the January issue. The system response time is set by the filtering at pin 3 of IC2, and is sufficiently fast to give a response with the stronger meteor-scatter signals.

Finally, a word of warning. Some Russian transmissions start at 4 a.m. our time. I can testify that SpE reception does occur at this time!

# A Question of Black Level 

## Malcolm Burrell

Black level is quite a useful test signal: it's also useful for video purposes. When making a programme, presentation comes a close second to the actual content. Adding a little black level between programme segments improves the editing, and a minute of black at the start of the tape, where most of the wear occurs, removes picture dropouts here. Thus when the cassette is inserted dropout effects should clear before the programme begins. In addition, rolling noise bars at the start are not seen. Tape noise and dropouts tend to be more easily seen on a plain blank raster, enabling the tape quality to be more easily assessed.

Black level can be obtained in several ways. One source is a camera with the lens capped. Another is a broadcast transmission, though you will be lucky to find a transmitter broadcasting black when you want it - the days are passed when transmitter line-up consisted of perhaps an hour of black-level and tone signal. Yet another source is a workshop pattern generator. Some don't give true black level without modification, so you might have to make do with a red or blue raster with the colour removed - some VCRs have a monochrome/colour switch.

## Black Level Generator

In the interests of convenience however I decided to make use of a Ferranti ZNA234 chip mounted in a compact case. This is basically a pattern generator i.c. that also provides fully interlaced sync pulses. It might seem a little wasteful, but only the mixed sync output is used. The 5 V supply required is obtained from a battery via a 78 M 05 UC stabilizer - the chip likes about 135 mA , which is more than mini-regulators can handle. The prototype uses a $2 \cdot 5 \mathrm{MHz}$ crystal, though this can be replaced with a single fixed capacitor of about 15 pF if you're not too concerned about stability.

The mixed sync output from the ZiNA234 drives a couple of gates in a TTL chip. The gate outputs are linked together via $27 \Omega$ resistors to give a $75 \Omega$ output. Connect the signal to the VCR via a short screened lead terminated
with a u.h.f., BNC or phono plug as required. You can thus record black level at any desired point by connecting the output to the video input socket and switching the machine to "aux" or "camera".

The unit can be easily constructed on Veroboard and accommodated in a small Vero box near the machine. Battery life will not be very long, but for short periods of use this is not very important. An LED indicator is useful next to the on/off switch.

You can't fade to black of course when using this blacklevel generator - it's much more of a cut to black level. Suppose however that you wanted to keep a selection of news items. You could judge the point at which the first is faded down, then add your black level for anything from three to ten seconds. You then edit in the next item as it is faded up.

This black-level facility is particularly useful with a VCR that doesn't have back-wind or insert-edit facilities, since most of the disturbance that arises will occur during black and thus be much less objectionable. It also makes a much cleaner distinction between the different items on the tape. The ZNA234 costs about $£ 8 \cdot 50$, so the total cost of construction need not be more than $£ 12$ or so. If you want to use the chip's other patterns, refer to the TV pattern generator article in the October 1981 issue.

Overseas readers with 525 -line signals can use this chip by connecting pin 2 to chassis instead of to 5 V and employing a $2 \cdot 52 \mathrm{MHz}$ crystal.


Fig. 1: Black level generator circuit.

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## FERGUSON 3V00

When playing back a recording the picture continuously moves from left to right by about three eights of an inch. The fault is more noticeable when the picture subject is stationary, and doesn't occur with prerecorded tapes.
It seems that the head speed is varying on record only. If the variation is erratic and random, suspect a dirty record/playback switch. This is the longer one of the two slide switches on the left-hand audio/servo board. Try cleaning it first. If the hunting is rhythmic and regular, the two head servo presets (discriminator and drum sample) probably need adjustment. Failing this the drum motor may be defective - check the ripple as described in the manual.

## SABA F CHASSIS

There is slight cramping, about a quarter of an inch deep, across the centre of the picture. It's only apparent when the camera shot is moving up or down and when the credits roll.
This can be an awkward problem and is due to nonlinearity in the field oscillator circuit. Check by substitution capacitors C691 ( $0 \cdot 1 \mu \mathrm{~F}$ ), C692 $(0.0022 \mu \mathrm{~F})$ and C693 $(0.022 \mu \mathrm{~F})$. They go slightly leaky and for some reason cause distortion at the centre of the field sawtooth waveform at pin 1 (triode anode) of the PCL805.

## GEC HYBRID COLOUR CHASSIS

The flywheel sync diodes used in this chassis share a common encapsulation, but the FSY41A doesn't seem to be available any more. Any ideas for substitutes?
We've successfully used a pair of 1N914 or ITT2002 diodes in this position. A pair of 1N4148 or BY206 diodes should work just as well.

## RANK A823A CHASSIS

The line output transistors have been replaced and I now want to balance the line output stage (6L4/5). It seems however that the balancing network has been removed. How do I monitor the adjustment?

6R6 and 6C5 were omitted in later production. To balance the stage, adjust the coils for minimum picture width: a "null" should be found, and this corresponds with correct balance.

## ITT CVC25 CHASSIS

The problem with this set is vision interference - it takes the form of streaking horizontal lines on all channels. Removing the aerial input leaves a snowy raster with mushy sound: the interference remains but is not as bad.

The cause could be in the field timebase as there has been lack of height with top foldover on one or two occasions, though it lasted for only two-three seconds.
First check the connections between the printed panel and the mains switch - dry-joints can occur here. Then check the mains filter capacitor: replace this if it's of the cylindrical yellow type, as these are given to internal arcing. If the fault persists and no bad joints can be found around the power supply panel, in particular around the surge limiter resistors R76/7 and the BY133 h.t. rectifiers, concentrate on the field timebase. If the problem lies here it's more likely to be with the output transistor pair T8/9 than in the pluggable module.

## THORN 3000 CHASSIS

There's a good, clean monochrome picture but this becomes increasingly noisy as the colour is turned up. At the correct colour control setting the noise gives the impression of a poor level signal from the aerial, but the signal strength here is first class.

If the colour is locked, check transistor VT110 and the a.c.c. smoothing capacitor $\mathrm{C} 173(2 \cdot 5 \mu \mathrm{~F})$ on the i.f. panel. If necessary, check the chroma amplifier transistor VT309 and the blanking diodes W316/7 on the decoder panel.

## PHILIPS G11 CHASSIS

No sound or raster led to a check on the h.t. fuse which had blown. Tests in the line timebase failed to reveal anything amiss, but the only way of preventing fuse failure is to disconnect the h.t. supply to the line timebase. The h.t. is 180 V instead of 153 V .

The excessive h.t. is the cause of the problem and is likely to be due to failure of the BD201 active smoothing transistor. The 27 V zener diode D4021 usually fails as well. A common cause of this situation is the h.t. reservoir capacitor C4029 which develops loose internal connections at the rivets. Try gentle mains application via a 110 V transformer at first to save fuse blowing - a 100 W lamp will do if you don't have a transformer.

## RANK T22 CHASSIS

There's no colour on this set except when the colour-killer override link is connected. With the set slightly off tune there are colour patches on the screen. Replacing the three decoder i.c.s has made no difference.
We've known electrolytic capacitors cause this sort of thing in the T130A decoder. Check C83 ( $3 \cdot 3 \mu \mathrm{~F}$ ) and C85 ( $1 \mu \mathrm{~F}$ ) by substitution and $\mathrm{C} 84(0 \cdot 22 \mu \mathrm{~F}$ ) by replacing it with a polyester type if you find it's a tantalum electrolytic. If necessary check $\mathrm{C} 87(4 \cdot 7 \mu \mathrm{~F})$ as well.

## THORN 8500 CHASSIS

The problem is vertical black lines, about one eighth to a quarter of an inch wide, at roughly quarter inch intervals right across the screen: they are not straight but curve with the edge lines of objects on the screen. Neither changing channels nor cleaning the tuner contacts improves matters.
Check the condition and adjustment of the set tuner gain control R102. If this is o.k., suspect the tuner - we've known this effect to be caused by a form of tuner instability.

## GEC C2110 SERIES

There's poor field linearity, with bottom cramping and a white line that moves rapidly up and down the bottom half of the screen. Field collapse sometimes occurs for a few seconds, then a full, normal picture appears for a while before the linearity faults return.

These sets are prone to problems due to poor earthing of the field timebase panel (at PL28/1). You can check this by using a screwdriver to link the panel's earth print to the metal chassis. If all's well here and capacitors C457/8 (field charging) and C462 (bootstrap) are o.k., the field output transistors are probably faulty.


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Each month we provide an interesting case of television senvicing to exercise your ingenuity. These are not trick questions but are based on actual practical faults.

Good service costs money in the TV trade - the cost of service attention by a reputable and well organised firm can be considerable. In our part of the world (and most, we suspect) there are several rather shady operators who offer their services at very low rates and whose attention can be inexpert to say the least. From time to time we encounter the results of their efforts when a set owner, in desperation, brings us a set to sort out. Similar situations have been described in these pages by other contributors: here's one of our botch stories.

The set's owner, a rather shady character himself, brought in a dog-eared and frayed Tandberg 22in. TV set. He then went back to his van and brought out the back cover - always a bad sign. Without removing his handrolled cigarette from his mouth, he hissed in a conspiratorial whisper "give us an estimate mate, colour's all wrong". We switched the set on there and then so that if the tube was flat he could take it back with him. A quick check showed that all three colours were present, but the picture looked like a colour snapshot negative! We promised to contact him when we'd got to the bottom of the fault.

On the bench we started by turning the colour right down to check the grey scale tracking. It was reasonable, though marred by a degree of impurity on one side. We made a mental note to deal with that later. The set was then tuned to the local Channel 4 test pattern, and on advancing the colour control setting a strange sight met our eyes! In order, from left to right, the colour bars were magenta, yellow, red, cyan, blue and green. Someone had swapped the drive leads to the tube - but inspection proved otherwise. The leads were correctly positioned on the tube base panel and at the RGB output stages.

Tongue in cheek we interchanged the leads -R drive to the green cathode, $G$ drive to blue and $B$ drive to red. This produced reasonably good colour within the tracking and purity constraints previously mentioned. Suspecting some diabolical trickery in the decoder or RGB output stages we carefully examined the panel, but could see nothing amiss on either side.

With full colour bars still displayed, we made an oscilloscope check on the RGB drives. They were correct, with the right $\mathrm{R}, \mathrm{G}$ and B primary colour waveforms coming from the right amplifiers! Back to the c.r.t. base panel. We could see no modifications or wiring alterations, and to confirm that we were not becoming unhinged we checked at the appropriate tube pins themselves - nos. 2, 6 and 11 for $R, G$ and $B$ respectively with this A56-120X tube. The waveforms were correct.

When the man returned for his set he had a considerable bill to pay, mostly down to time rather than components. It had taken us half an hour to find a means of refitting the set's back cover, and rather longer to sus out the cause of the colour problem. You needn't concern yourselves with the back screws, just the wrong colour effect: answer next month!

## ANSWER TO TEST CASE 254 - page 217 last month -

The problem described last month was no field scan in a Doric colour portable fitted with the Rediffusion Mk. 5 chassis and our attempts at diagnosis, having twice convinced ourselves that the field oscillator/driver chip was faulty. The virtually zero d.c. voltage at pin 3 of the chip should have pointed us in the right direction, for if Q502 was not switching on - and it wasn't, without any drive there should have been a fairly high voltage at its collector and thus at pin 3 of the i.c., due to the conduction of Q501.

Q501 was not conducting however, for the very good reason that its base bias resistor (Q502's load resistor) was open-circuit. It looked all right, but there was 73 V at one end and nothing at the other. The design of the field timebase chip is such that with zero feedback at pin 3 it shuts down, which is what was happening. We must confess that we were happier with PCL805s and the like sometimes you got a dud spot on the height control, or the valve got broken...


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| . MC 1310P | 120 | TAA 550 | 0.39 | UPC 574 | 0.55 | 8Y133 |  |  |  |  |  |  |  |  |  |
| MC 1327 P | 1.10 | TAA 691 | 1.95 | UPC 57562 | 1.99 |  | 1.35 | ${ }^{\text {BCI }}$ | 0.09 | ${ }^{\text {Buazav }}$ | 1.70 | SWITCHES |  | COMP |  |
| MC ${ }^{1330}$ | 1.50 | TAA 700 | 1.65 | UPC 577 | 3.50 | BY184 | 0.40 | ${ }_{\text {BC220 }}$ | 0.09 |  |  |  |  |  |  |
| MC 1349 P | 1.10 | TBA 120A | 0.75 |  | 1.19 |  | 0.69 | BC2121 | 0.09 | E122 |  |  |  |  |  |
| MC 1351 | 120 | tBA 12as | 0.75 | 7419 | 0.95 |  | 0.60 | BC213L | 0.11 | ${ }^{\text {F33055 }}$ | 0.55 | Pull | 0.00 | Decca 10 |  |
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| ML 2368 | 3.00 | tba l20SA | 0.75 | UPC 1018 C | 1.19 | ${ }^{\text {BY}}$ | 0.21 | BC250 | 0.11 |  |  |  |  |  |  |
|  | 1.00 | tBA 1201 | 0.75 | PC 1 | 0.63 |  | 0.31 |  | . 11 |  |  |  |  |  |  |
|  | 3.00 | tBA 231 | 1.55 | UPC 10 | 0.63 | 8 P 295 | 0.25 |  | 0.10 | R202 |  |  |  |  |  |
| SAA 1024 | 2.40 | tBA | 1.70 | UPC 10 | 3.70 | $8 \mathrm{Br26}$ | 0.25 |  | . 11 | R203 | ${ }^{\text {. }} 90$ | 1500 ootary |  |  |  |
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| SAA 130 | 3 | TBA 510 | . 20 | 103H2 | 2.00 | 81810 | 0.21 |  | 33 | 223 |  |  |  | CRYSTAL |  |
| ${ }_{\text {SAS }}$ S50] | \% 10 |  | 1.50 | PCC 1035 S | 250 |  | 0.50 | BC388 | 0.13 | ${ }_{\text {R2322 }}$ | ${ }_{0.70}$ | Philips 611 | 1612 |  |  |
|  | \% 0 |  | . 1.5 | Prc 1035 | 2.50 | 8xpl1350 | 0.50 | BC327 | 0.11 | ${ }_{\text {R2354 }}$ | 0.78 | Remote | 1.30 | ${ }_{8.860 .4}^{4.4}$ | \% |
| ${ }_{\text {Sas }}$ SE8 | 1.00 | ${ }_{\text {TiAA 5 50 }}$ | . 10 | UPC | 1.28 | Bx71.60 | 0.09 | BC328 | 0.13 |  |  | Thern 1615 | wit |  |  |
| SAS 670 | 230 | TBA 500C | 1.90 | UPC 1022 C | 2.60 | ${ }_{\text {BRILOO }}$ | 0.35 | ${ }_{81}^{8 C 3}$ | 0.11 | ${ }_{\text {R2461 }}^{\text {R241 }}$ | 2.00 |  |  | Minitur | 1.75 |
| SL 9018 | 4.50 | TBA 570 | 1.10 | UPC 1033C | 2.60 | OA47 | 0.10 | BC350 | 0.12 | ${ }^{82477}$ | 0.78 | ${ }_{35000}^{301}$ |  |  |  |
| SL. 9178 | ${ }^{6} 2.20$ | TBA 750 | 1.50 | UPC 1156\% | 2.10 | OAsO | 0.08 | 析 | 0.30 | ${ }_{\text {R2501 }}^{\text {R2501 }}$ |  | Veri Gun Su | witch 0.5 | VIDEO |  |
| SN 1168824 N | 2.50 | TBA 800 | 1.10 | UPC 1158 H | 0.78 | OA91 | 0.00 | ${ }^{\text {BCa }}$ | 0.11 | ${ }^{\text {cherese }}$ |  | (A1) |  | FILMS |  |
| SN 23844 | 2.25 | tiA 820 | 1.20 | UPC 11133 | 0.9 | ${ }_{\text {INSO }}$ | 0.08 | ${ }_{\text {che }}^{\text {BCas60 }}$ | 0.52 |  |  | Horiz |  |  |  |
| SN 51108 AN | 1.95 | TBA 890 | 3.50 | UPC 116772 | 1.58 | 114001 | 0.05 | BC462 | 0.53 | 90055 |  |  |  |  |  |
| SN 76001 AN | 1.60 | TBA 920 | 2.00 | UPC 1168 C | 2.75 | N4002 | 0.05 | CC46 | 0.53 |  |  |  |  |  |  |
| SN 76003 | 220 | TBA 950 | 2.65 | UPC 11700 | 1.75 | 114003 | 0.05 | C55 | 0.13 |  |  |  |  |  |  |
| SN 76013 N | 2.50 | TBA 990 | 1.60 | UPC 1717c | 1.62 | Na04 | 0.06 | ${ }^{8}$ | 0.13 |  |  | DELAY L | INES |  |  |
| SN 76013ND | 2.50 | ${ }^{\text {TBA }} 1440$ | 2.70 | jPC 1173 Cl | 2.14 | N403 | 0.01 | ${ }^{8 C 5}$ | 0.10 | ${ }_{\text {HeP3IC }}$ |  |  |  | FILTERS |  |
| SN 18023 N | 220 | ${ }_{\text {TBA } 1414}$ | 2.70 |  | 2.13 | IN4006 | 0.01 | ${ }_{\text {BC548 }}$ | 0.09 | ${ }_{\text {T1P32 }}$ | 0.34 | 1.700 |  |  |  |
| SN 761312 | 1.35 | TCa 270 s | 1.90 | UPC 117 m |  |  | 0.07 | ${ }^{\text {cc5 }}$ | 0.10 | T1P32 |  |  |  |  |  |
|  | 1.120 | Tca 27050 | ${ }_{2} 20$ | UPC 11180 C | 2,14 300 | 1 Na | 0.03 |  | 0.10 |  |  | Luminanco | Imatches | sW154 | 220 <br> 220 |
| SN 7653 | 2.00 | TCA 650 | 2.40 | UPC 1181н3 | 1.62 | INS | 0.15 | ${ }^{80131}$ | 0.30 | ${ }_{\text {TiPa }}$ | ${ }_{0} .58$ |  |  |  |  |
|  | 0.29 |  | 2.80 | UPC 118 | 2.70 | IN506 | 0.11 | ${ }_{80135}$ | 0.30 | ${ }^{\text {TPPP2AA }}$ | 0.45 |  |  | SUNDRIE |  |
| SN 76546 | 1.00 | TCA 740 | 2.90 | UPC 11 | 2.30 | INS |  | 1336 |  | T1P30 |  |  |  |  |  |
|  | 0.90 | 800 | 2.00 | UPC 118542 | 3.50 | Y969 | 0.87 | ${ }^{8139}$ | 0.33 | TITP3 | ${ }_{0}$ | TUN |  |  |  |
| SN 6866 N | 1.10 | tca 830 S | 1.60 | UPC 11886 | 0.10 |  |  |  | 0.60 |  |  | PRESETS |  | Biock 2 A |  |
| SN 7 T 706 | li.70 | TCA 4500A | 1.80 | UPC ${ }_{\text {U }} 11188 \mathrm{~g}$ |  | ZENERS |  |  | 0.70 | ${ }_{\text {T13P2A }}$ |  |  |  |  |  |
| TA 003 BP | 3.50 | TDA 1003 | 2.00 |  | 2.00 | 400 M |  | B02 | 0.40 | ${ }_{\text {T1P33 }}$ | ${ }_{0}^{0.75}$ |  |  |  |  |
| TA 70 gesp | 3.30 | TDA 1004 A | 220 | UPC 11919 | 1.70 | ${ }_{12 \mathrm{Na}}^{20}$ | 0.10 | B02 | 0.35 | TTP41A | ${ }^{0} 4.45$ |  |  | biad |  |
| TA 7072 zr | 230 | TAA | 2.50 | 190 | 1.70 |  | 0.18 |  |  |  |  |  |  |  |  |
| TA 074P | 3 | To | 2 | UPC 130\% | 1.97 | 100V to 200 V | 0.19 | ${ }_{\text {B0238 }}$ | 0.38 | T1P42C |  |  |  |  |  |
| TA 70 | 303 | TOA | 2 |  | 1.16 | 15 wat |  | 8023 | 0.12 | TiP100 |  | EHT LE |  | GEC |  |
| TA 7098P | 2.40 | TDA 1327 | 1.70 | UPC 1208 C | 2.00 | BZVICCI2R |  | ${ }^{\text {B02338 }}$ | 0.65 | T1P2955 | 0.74 |  |  | ${ }_{\text {cosen }}^{3500}$ cuns |  |
| TA 7osp | 230 | TDA 1412 | 1.00 |  | 4.00 | ${ }^{\text {BZVIICCOR }}$ | 1.18 | 8046 | 0.23 |  |  | diode |  |  |  |
| TA | 2.4 | ToA 2002 | 1.90 |  | 1.34 |  |  | ${ }^{80595}$ | 0.55 | ${ }_{\text {Tis91 }}$ |  | EHT Cap |  | Hank Tu |  |
| TA 7120 P | 1.61 | TOA 2522 | 2.80 |  | 2.10 | BRIDGE |  | ${ }^{8055}$ | 0.55 | 2N3055 | 0.73 | 10 M Pack | of EHT | Cams |  |
| TA 7129AP | 320 | TDA 2523 | 2.20 | UPC 1216 V | 1.99 | 840 |  |  | 0.50 | 28835 |  |  |  | 5yds PvC |  |
| TA 1330 | 1.30 | ToA 2330 | 2.20 |  | 3.59 |  | 0.50 |  |  |  |  |  |  |  |  |
| TA 71375 | 2.00 | TAA 2332 | 2 |  | ${ }_{2} .300$ |  | 0.70 | ${ }^{3 F 127}$ | 0.30 | THYRIST |  |  | DIS | COUN |  |
| TA 71468 | 2,75 2.00 | TOA 1365 | 5.27 | UPC 1223 | ${ }_{3.70}$ | wos |  | ${ }^{\text {BFIB0 }}$ |  | KONIG 15/8 | Universal |  |  |  |  |
| TA 715P | 3.20 | ToA 2541 | 2.30 | UPC 1225 | 3.00 | TRANSIS |  |  | 0.27 | KONIG 15/8 | Universal | sal $\quad 2.47$ | ORD | ERS OV |  |
| TA 717 T | 320 | ToA 2560 | 2.05 | UPC 1226 | 2.96 |  |  |  | 0.23 | BR101 |  |  | £50 | IN VA |  |
| TA $1717 \mathrm{~S}^{\text {ap }}$ | 3.00 | TOA 2581 | 2.10 1.30 | UPC 1227 | 2000 | ${ }^{\text {ACL128 }}$ | 0.30 0.30 | ${ }_{\text {coser }}$ | 0.10 |  |  | ${ }_{0.56}^{0.62}$ |  | 5\% |  |
| TA 7193P | 5.20 | TDA 2582 | 1.80 | UPC 1230H | 3.60 | AC153\% | 0.36 | SFes | 0.11 | BYY49 |  | 0.65 |  |  |  |
| ta 7203P | 330 | TOA 2593 | 2.70 |  | 1.90 | Ac176 | 0.30 | ${ }_{8 F 197}$ | 0.13 | ${ }^{\text {BrYY5 }}$ |  |  | ORD | ERS OV |  |
|  | 3.16 2 | TDA 2661 A | 1.15 | UPC 1245 | 2.20 | ${ }_{\text {AC }}^{\text {Al187 }}$ | 0.30 |  | 0.16 | ${ }^{\text {BrIT50 }}$ |  | ${ }_{0} .30$ | £120 | IN VAL |  |
| TA 7205P | 1.40 | TDA 2640 | 1.80 | UPC 1350 C | 4.50 | AC188 | 0.30 | FF20 | 0.31 | ${ }^{\text {BTITOG }}$ |  | ${ }^{55}$ |  | \% |  |
| TA 2 208P | 2.70 | tod 2633 | 2.10 |  | 2.80 | A0149 | 0.88 | ${ }^{2} 240$ | 0.16 | ${ }^{\text {Br1166 }}$ |  |  |  |  |  |
| ${ }_{\text {TA }}^{\text {TA } 21208}$ | 5.50 | ToA 2680 | 2.10 | UPCC 13556 | 3.00 | A0162 | 0.05 | 8245 | 0.30 | ${ }_{\text {BTI20 }}$ |  | 2.00 |  | Va |  |
| TA 723 P | 3.50 | TOA 3560 | 5.10 | UPC 1350 C | 3.50 | AF139 | 0.48 | BF258 | 0.27 | DEC1 |  | . 70 |  | ding VAT |  |
| TA 7227P | 5.10 | TDA 3561 | 6.50 |  | . 70 |  | 0.60 | ${ }^{\text {B2F299}}$ | . 32 |  |  |  |  |  |  |
| TA filiop |  |  |  |  | 3.50 | AU113 | 2.60 | ${ }^{\text {br273 }}$ | 0.13 0.0 | OT121 |  | 1.55 1.55 | CRTs | EXCLI |  |
| TA 313 P | 2.80 | TDA | 2.10 | UPC 13 | 5.00 |  |  |  |  |  |  |  |  |  |  |



| AA117 | 0.050 | BC157 | 0.055 | 80222 | 0.310 | BU108 | 1000 | 0 C 72 | 0.500 | TIP29C | 0.250 | 3N. 128 | 0.550 | 2 SA473 | 0370 | 28 pin | 0.200 | PCLE2 0.830 | LM723 0320 | ED 5m |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AA119 | 0.050 | BC159 | 0.055 | 80225 | 0310 | BU110 | 1.100 | OC200 | 1800 | TIP30 | 0.180 | 3N. 143 | 0.850 | 2SB54 | 0250 | 40 | 0250 | PCLA 0.500 | LM741 Dil 0.150 |  |
| AAY32 | 0.050 | BC182 | 0.060 | B0232 | 0.310 | BUl11 | 1400 | OCP71 | 1.000 | TIP31A | 0210 |  |  |  |  |  |  |  | LM741 Di 0.150 | 0 |
| AC107 | 0.230 | BC182L | 0.080 | B0237 | 0210 | BU128 | 0.700 | ORP12 | 1.000 | T1P32 | 0210 | IN. 914 | 0.020 | $2 \mathrm{SB77}$ |  |  |  | PCL85 0.550 | LM741 | 5 5mm |
| AC126 | 0.170 | BC183 | 0.060 | B0238 | 0240 | BU204 | 0.750 | ORP60 | 1.000 | TIP32A | 0210 | IN.4001 | 0.010 | 258337 | 1200 | VALVES |  | PCL86 0.550 | Mot 0,450 | GREEN 0.100 |
| AC127 | 0.150 | BC183 | 0.060 | B0430 | 0280 | BU205 | 0.700 | ORP61 | 1.000 | TIP33 | 0.500 | IN. 4002 | 0.040 | 2SB405 | 0220 | DY87 | 0.50 | PCL805 0.550 | LM3900 0.250 |  |
| AC128 | 0.150 | BC184 | 0.060 | B0437 | 0280 | BU208 | 0.750 |  |  | TIP34 | 0.500 | IN. 4003 | 0.040 | 2SC480 | 0210 | DYBCP | 0450 | PFL200 0850 | NE555 0.150 | BRIDE |
| AC128K | 0.280 | BC184L | 0.060 | 80535 | 0300 | BU208A | 0.800 | R2008B | 0.800 | TlP41A | 0.220 | IN. 4004 | 0.040 | 2SC495 | 0.800 | ECC82 | OACO | PL36 0800 | NE556 0.400 | RECTIFERS |
| AC141K | 0.230 | BC212 | 0.050 | BD536 | 0380 | BU2080 | 1200 |  |  | T1P41C | 0.250 | IN. 4005 | 0.040 | 2SC733 | 0.400 | ECCEs | 0430 | PL504 O.950 | BYX55 | 1A50V 0.100 |
| AC142K | 0.220 | BC212L | 0.060 | 80537 | 0400 | BU326 | 0.850 | SAS560 | 1,100 | TIP42A | 0.220 | IN. 4008 | 0.010 | 2SC1161 |  | ECC84 | 0400 | PL508 1.900 | $350 \quad 0300$ | IA 100V 0.180 |
| AC153K | 0.230 | BC213 | 0.060 | 80538 | 0400 | BU406 | 0.850 | SAS570 | 1.100 | TIP42C | 0250 | IN. 4007 | 0.050 | 2SC1172Y | 1.100 | ECCOA | 0400 | Pl508 1.500 | $350 \quad 0300$ | 1A 100V 0.160 |
| AC176 | 0.180 | BC213L | 0.080 | $80 \times 32$ | 1.000 | BU407 | 0.750 | SN76003 | 1.400 | T1P47 | 0.400 | IN. 4148 | 0.020 | 2SC1172Y | 1.500 | ECC8S | 0400 | PY81 0.700 | BYX55/ | 1A200V 0.150 |
| AC176K | 0.200 | BC214 | 0.060 | BDX65 | 0800 | BU408 | 1800 | SN76013 | 1400 | TIP48 | 0400 | IN. 5400 | 0.050 | 2SC1279 | 0240 | ECH81 | 0490 | PY88 0.480 | 60000300 | 1 A 400 V 0.210 |
| AC187 | 0.150 | BC214L | 0.060 | BF180 | 0.160 | BU500 | 1.100 | SN76023 | 1400 | TIP49 | 0400 | IN. 540 | 0.100 | 2SC1308 | 1.000 | ECH84 | 0.520 | PY500A 1.600 | 81×55 | 1A 600 V 0.230 |
| AC187K | 0.200 | BC237 | 0.070 | BF181 | 0.180 | BU526 | 0800 | SN76033N | 1.500 | TIP10 | 0270 | IN. 5402 | 0.100 | 2SC1307 | 1.000 | ECLSO | 0.570 |  | 6000300 | 1A B00V 0230 |
| AC188 | 0.170 | BC238 | 0.070 | BF1E3 | 0200 | BY126 | 0.060 | SN76110N | 0.700 | TIP112 | 0.540 | IN. 5403 | 0.110 | 2SC1520 | 0250 | ECLB2 | 0.590 |  | BYX55/ | 2A100N 0350 |
| AC188K | 0.230 | BC300 | 0.160 | BF194 | 0.200 | BY127 | 0.080 | SN76115 | 0.700 | TIP115 | 0250 | IN. 5404 | 0.110 | 2SC1969 | 1300 | ECL84 | 0.570 | ZENERS | 80000320 | 2A 200V 0300 |
| ACY18 | 0.480 | BC301 | 0.180 | BF195 | 0200 | BY133 | 0.030 | SN76226 | 0.500 | TIP117 | 0.560 | IN. 5405 | 0.120 0.130 | 2SC2029 | 1200 | ECL85 | 050 | 400MV | 80000300 | 2AN400V 0120 |
| ACY19 | 0.480 | BC302 | 0.180 | BF194 | 0.050 | BY164 | 0220 | SN76227 | 0.800 | T1P120 | 0.430 | N. 540 | 0.130 | 2Sc2079 | 1200 | ECL85 | 0.570 | BZY88 Range | BYX70 | 2ANGONV 0132 |
| AD142 | 0.600 | BC303 | 0.180 | BF195 | 0.050 | BY176 | 0850 | SN722 | 0.0 | TIP121 | 0.460 | IN. 54007 | 0. | 2SC2078 | 120 | EC | 0 | 2V7 to 39V0.050 | 5000290 | 2AN600 0.510 |
| AD149 | 0.450 | BC327 | 0.060 | BF196 | 0.050 | BY179 | 0350 | T28000 | 0.520 | TIP 122 | 0470 | IN. 5408 | 0.130 | 2SC2122A | 2000 | Ef80 | 0310 | 2V7 | BYX70 | 2ABSOV 0.500 |
| AD161 | 0.220 | BC382 | 0.060 | BF199 | 0.060 | BY182 | 0320 | TAG06-60 | 0820 | TIP125 | 0470 |  |  | 2SC2952 | 0.270 | EFP5 | 0.340 |  | 5000310 | 3A200N 0.5e0 |
| AD162 | 0.220 | BC337 | 0.060 | BF200 | 0.160 | BY184 | 0220 | TAG521- |  | TIP126 | 0.560 |  |  | 2SD234 | 0.370 | Ef89 | 0.330 | - | BYX70 | 3A 400V 0.800 |
| AF124 | 0250 | BC328 | 0.060 | BF257 | 0.180 | BY187 | 0320 | 200 | 0.720 | TIP127 | 0.560 | REGU | Ons | 2SK135 | 4.000 | EF183 | 0450 | 2 V 7 to 39V0.12 | B00 03 |  |
| AF125 | 0250 | BC557 | 0.060 | BF258 | 0.180 | BYIs6 | 0200 | TAG4443 | 0.760 | TIP2955 | 0340 | 7805 | 0350 | 2 N 135 | 4.00 |  |  |  | 800 | 3A 600 |
| AF126 | 0.250 | BCY32 | 1.500 | BF259 | 0.180 | BY206 | 0.110 | TAG4444 | 0.760 | TIP3054 | 0380 | 7812 | 0350 | 337 | 1.500 | EF184 | 0.5 |  | 84x71/ | 6 A 200 V 1.000 |
| AF127 | 0250 | ВС「33 | 1.500 | 8F336 | 0.200 | BY207 | 0.110 | TAA550 | 0.150 | TIP3055 | 0340 | 715 | 0350 | TA7205 | 1.500 | El34 | 1.500 | memories | 60000.800 | 6A 400V 0.800 |
| AF139 | 0.220 | $8{ }^{8}$ | 1.500 | BF337 | 0200 | Br223 | 0.720 | TBAI2OS | 0.950 | TIS61 | 0.150 | 818 | 350 | UPC575 | 1.000 | EY86 | 0310 | 21140.750 |  | 25A 100V 1.600 |
| AF239 | 0.20 | BCY42 | 0.200 | 8F338 | 0200 | BYx10 | 0.150 | TBA395 | 0.600 | TIS90 | 0.150 | 7624 | 0350 |  |  | EY87 | 0310 | 27162300 |  |  |
| AL112 | 0.700 0800 | BCY56 BCY70 c-771 | 0.160 0.160 | 8 8362 | 0300 | CA270 | 0400 | T8A396 | 0.600 | TIS91 | 0.180 | 7905 | 0350 | cs |  | PC87 | 1000 | $2532 \quad 2.500$ |  | EIECTHOLYTIC |
| ASZ15 | 1.000 | BCY71 | 0.160 | 8F458 | 0.190 | CA3086 | 0250 | TBA530 | 0.750 |  |  | 7912 | 0800 | SOCXETS |  | PCC85 | 0.420 | $2732 \quad 2.500$ |  | 4700UF-16V |
| ASZ17 | 1.000 | BCY72 | 0.160 | BF459 | 0.150 | CA3089 | 1.500 | TBA540 | 0.750 | 2N. 2904 | 0.200 |  | 0.400 | 8 pin | 0.060 | PCF\% | 0.580 | 27645 | LED 3 mm | CAN 0.200 |
| AU106 | 1.000 | 80115 | 0250 | BFX29 | 0200 | CA3240 | 0.900 | TBA560 | 0.700 | 2N. 2905 | 0200 | 7918 | 0400 | 14 pin | 0.080 | PCF200 | 1350 | 411600.750 |  |  |
| AU110 | 1.100 | BD124P | 0.500 | BPX84 | 0200 | C106D | 0280 | TBA800 | 0350 | 2N. 2906 | 0.180 | 75 | 0.400 | 16 pin | 0.090 | PCF801 | 1.100 | 61183.000 |  | as |
| AY102 | 1800 | 80124 | 1.100 | BFX85 | 0200 | MC1327 | 0.700 | TBA810S | 0.600 | 2N. 2907 | 0.180 | 78.05 | 0230 | 18 pin | 0.120 | PCF802 | 0.570 | LM324 0300 | LED 3mm | LP1195 |
| AY106 | 1800 | BD128 | 0350 | BPX87 | 0.150 | MJ2500 | 1.000 | TBAs20 | 0.750 | 2N. 2926 | 0.080 | 78L12 | 0.290 | 20 pin | 0.140 | PCF806 | 1.150 | LM330 0.800 | GREEN | 2250 |
|  |  | BD131 | 0250 | 8FX88 | 0.150 | MJ2501 | 1.100 | TBAgzo | 0.800 | 2N. 3019 | 0230 | $78 L 15$ | 0.290 | 22 pin | 0.150 | PCH200 | 1.000 | LM381 |  |  |
| BA145 | 0.100 | 8D132 | 0250 | BFY50 | 0.140 | MJ2955 | 0.550 | TBAS50 | 0800 | 2N. 3053 | 0.180 | 74.18 | 0200 | 22 pin | 0.180 | PCI81 | 1.00 | LM709 | 5mm |  |
| BA148 | 0.100 | BD135 | 0200 | BFY51 | 0.140 | M 33000 | 1.150 | TBA990 | 0800 | 2N. 3054 | 0350 | 74124 | 0280 | 24 pin | 0.180 | PCL81 | 0.540 | LM709 Dil 030 | 0.050 |  |
| BA154 | 0.050 | 8 B 136 | 0200 | BFY52 | 0.140 | M J3001 | 1,150 | TCA800 | 0 000 | 2N. 3055 | 0230 |  |  |  |  |  |  |  |  |  |
| BA157 | 0.120 | BD137 | 0200 | BFY56 | 0250 | MJE23A | 0.300 | TCA940 | 0850 | 2N.3055 | 0380 | LM309K | 1.000 | Please add 40p. P\&P and VAT at 15\%. Govt. Colleges, atc. |  |  |  |  |  |  |
| 88101 | 0.130 | ${ }^{8 D 138}$ | 0200 | BYF57 | 0250 | MJESOA | 0300 | TDA1170 | 0.500 | 2N3440 | 0 0.800 | LM317K | 2200 |  |  |  |  |  |  |  |
| B8103 | 0.160 | 8D139 | 0200 | BFY84 | 0250 | MJE340 | 0.250 | TDA1412 | 0.800 | 2N 3442 | 0.850 | LM31T | 1.800 | orders accepted. <br> Quotations given for Large Quantities. |  |  |  |  |  |  |
| B81058 | 0.120 | 8 B 140 | 0200 | BRI00 | 0.140 | MJE350 | 0800 | TDA2002 | 0800 | 2N3771 | 0850 | LM323K | 4.200 |  |  |  |  |  |  |  |
| 882058 | 0240 | 8 B 144 | 0.500 | 8SX19 | 0.150 | MJE520 | 0.300 | TDA2006 | 1.500 | 2N 3772 | 0.800 | LM723 | 0320 | Please allow 7 days for delivery |  |  |  |  |  |  |
| BC107 | 0.070 | 8 BD 50 | 0300 | BSX20 | 0.150 | MJE2965 | 0.500 | TDA2000 | 1400 | 2N3773 | 1.000 | 78HGKC | 5.700 |  |  |  |  |  |  |  |
| BC108 | 0.070 | BD157 | 0300 | 85×21 | 0.180 | M 7 | 0 | TDA2050 | 1400 | 2N. 4031 | 0.250 | 78H05KC | 5.200 |  |  |  |  |  |  |  |
| 8C108 BC115 | 0.070 | $8 \mathrm{8D} 58$ | 0380 | BSX26 | 0.160 | OA47 | 0.080 | TDA2522 | 0800 | 2N. 4036 | 0250 | 78GUIC | 1.900 |  |  |  |  |  |  |  |
| BC 115 $\mathrm{BC118}$ | 0.100 | ${ }^{8 D 168}$ | 03500 | 8SX29 | 0.180 | OASO | 0.010 | TDA2530 | 0800 | 2N. 4037 | 0250 | 79GUIC | 2.150 | GRANDATALTD. |  |  |  |  |  |  |
| BC118 BC140 | 0.110 | 88175 | 0300 | 82776 | 0.180 | A091 | 0.040 | TDA2532 | 0.750 | 2N. 4443 | 0.780 | 79HGKC | 6.700 |  |  |  |  |  |  |  |
| BC140 | 0.150 | 8017 | 0300 | BT106 | 0.500 | OA200 | 0.070 | TDA2540 | 0.700 | 2N. 4444 | 0.760 | Japanese |  |  |  |  |  |  |  |  |
| BC141 | 0.180 | B0179 | 0320 | BT109 | 0.900 | OA202 | 0.070 | TDA2580 | 0.300 | 2N. 5061 | 0200 |  |  |  |  |  |  |  |  |  |
| BC142 | 0.180 | BD181 | 0950 | BT116 | 0.200 | OC2B | 1.000 | TDA25s | 0.900 | 2N. 6294 | 0300 | TRAN8ESTORS |  |  |  |  |  |  |  |  |
| BC143 | 0.180 | 8D209 | 0330 | BT118 | 1.000 | OC29 | 0.800 | TDA2840 | 0.800 | 2N. 5298 | 0300 | 2SA73 | 0300 | WEMBLEY, MIDDLESEX, ENGLAND. |  |  |  |  |  |  |
| BC147 | 0.085 | 8D20 | 030 | BT120 | 1.000 | OC35 | 1.000 | TDA2680 | 0.700 | 2N.6108 | 0400 | 2SA104 | 0320 | Telephone: 01-904 2093 \& 904-1115/6. |  |  |  |  |  |  |
| BC148 | 0.050 | 3D203 | 0.420 | 8U104 | 1.000 | 0C45 | 0.500 | T\|P29 | 0.150 | 2N. 6107 | 0400 | 2SA198 | 0220 |  |  |  |  |  |  |  |
| BC148 | 0.055 | 8D204 | 0820 | BU105 | 0800 | 0C71 | 0.300 | TIP29A | 0.220 | 2N. 8109 | 0.400 | 2SA203 | 0300 | Telex No. 932885 (Sunmit) |  |  |  |  |  |  |



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