PRACTICAL ELECTRONICS MARCH 1982 75p

Use Your TRANSISTORS with the B PROJECTS in this issue!!













telectric Cost Monitor Plus... EMERGENCY LIGHT

Much more than just kits quite simply the best way to make music

The smart contemporary styling of the 'black-boxes' contain the easy-to-build advanced electronics that produce everything from high grade amplifcation to complex lighting effects. These units are among the finest of their kind available and combine, as do all the Powertran kits, constructional ingenuity with high grade performance capability.

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Each unit can, of course, perform its independent function - but it is compatible with its fellows (same cabinet sizes and the same quality and professional finish) to enable you to assemble an impressive bank of wholly controllable power.

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MPA 200 – a 100W amplifier with professional finish and

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CHROMATHEQUE 5000 - 5 channel lighting effect system. Sound to light, strobe to music level, random or sequential effects - each channel handles up to 500W yet minimal wiring is needed with our COMPLETE KIT £49.50 (+ VAT) single-board design.

ETI VOCODER - 14 channel for maximum versatility and high intelligibility; each channel has independent level control. 2 input amplifiers - for speech/external excitation - each with level control and tone control. The Vocoder is a superb machine capable of infinite variety of operation. Contstruction with our easy to follow, comprehensive builders' manual – is challenging yet within the scope of most enthusiasts. COMPLETE KIT £175 + VAT)

> SP2 200 - 2 channel 100W amplifier. Two of the rugged, reliable and economic amplifiers from the MPA200 are fed by separate power supplies from a common toroidal transformer. Fully finished metalwork, fibreglass PCBs, controls, wire - everything you need to make this powerful COMPLETE KIT £64.90 + VAT) unit.

> > DJ90 STEREOMIXER - a

versatile new mixer with 2 stereo inputs for magnetic cartridges, a stereo auxiliary input and mike input. Auto planning for fast or slow, slider controls, multi-mixing, ducking, interrupt, input modulation - everything yet still under £100! (Our console below shows the mixer neatly teamed with a Chromatheque and SP2 200) **COMPLETE KIT £97.50** + VAT



VOCODER

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VOLUME 18 No. 3 MARCH 1982

CONSTRUCTIONAL PROJECTS

TELECTRIC ELECTRICITY COST MONITOR Part 1 by Stephen Day BSc Eng MIEE	 22
Can provide instant and long-term cost monitoring of almost any electrical appliance	
EMERGENCY LIGHT by Stephen Ibbs	 28
Rechargeable, with automatic cut-in	
FREE TRANSISTOR PROJECTS	 32
Six circuits specially designed around your free transistors	
AUTO LIGHT, CAR BATTERY CONDITION INDICATOR, AF/RF GENERATOR,	
BENT WIRE GAME, DOUBLE SIREN, LIQUID ALARM	
INFRA-RED REMOTE CONTROL Part 2 by Bart Trepak	 40
Construction and applications	
PE RANGER BASE STATION Part 3 by Michael Tooley BA and David Whitfield MA MSc	44
Testing and alignment	
PE ROBOTS Part 5 by Richard Becker, Tim Orr and Richard Monkhouse	56
Resume, completion of hardware, and software prologue	
TV CAMERA Part 3 by Philip Gaffney	 62
Assembly and setting up	

GENERAL FEATURES

DIGITAL DESIGN TECHNIQUES Part 8 by 7	om Gas	kell BA	 	 	 48
Memories and Microelectronics					
MICROBUS by D.J.D.	1.10		 	 	 60
Use of micros in amateur radio					
SEMICONDUCTOR UPDATE by R. W. Coles			 		 66
Featuring LM1035 99000 Series 8X60					

NEWS AND COMMENT

EDITORIAL		 					1	7
NEWS & MARKET PLACE		 					1	8
Including Countdown					 	•••		-
INDUSTRY NOTEBOOK by Nexus		 			 		2	21
Reflections on the past year								
PATENTS REVIEW		 			 			64
New ideas from around the world								
SPACEWATCH by Frank W. Hyde							e	38
Extra-terrestrial activities chronicled	i i	 	•••	• •	 	•••		

OUR APRIL ISSUE WILL BE ON SALE FRIDAY, 12th MARCH 1982

(for details of contents see page 39)

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22n 12p; 33n, 47n, 68n 16p; 100n, 150n 20p; 220n 30p; 330n 42p; 470n 52p; 880n 60p; 1µF 68p; 2µ2, 4µ7 85p, 160V; 10nF, 12n, 100n 11p; 150n, 220n 17p; 330n, 470n 30p; 680n 38p; 1µF 42p; 1µ5 45p; 2µ2 48p; 4µ7 58p. 1000V; 1nF 17p; 10nF 30p; 15n 40p; 22n 38p; 33n 42p; 47n, 100n 50p.	BC108C 12 BD140 40 MJE340 54 TIP2955 60 2N365 15 2SC2091 85 BC109 10 BD144 198 MJE370 100 TIP3055 60 2N3614 199 2SC2314 85 BC109B 12 BD205 110 MJE371 100 TIP3055 60 2N3614 199 2SC2314 85
POLVESTER RADIAL LEAD CAPACITORS: 250V; 10n, 15n, 22n, 27n 6p; 33n, 47n, 68n, 100n 7p; 150n, 220n 10p; 330n, 470n 13p; 680n 19p; 1µ 23p; 1µ5 40p; 2µ2 46p.	BC109C 12 8D245 45 MJE520 95 TIS44/45 45 203663 16 2SC1679 190 BC117/8 20 BD378 70 MJE2955 99 TIS48 50 2N3702/3 10 2SD234 75 BC140/42 30 BD374 55 MJE2055 70 TIS48 50 2N3702/3 10 2SD234 75
ELECTROLYTIC CAPACITORS (Values in µF). 500V: 10 52p; 47 78p; 63V: 0.47, 1.0, 1.5, 2.2, 3.3, 8p; 4.7 9p; 6.8, 10 10p; 15, 22 12p; 33 15p; 47 12p; 100 19p; 1000 70p; 50V: 47 12p; 68 20p; 220 24p; 470 32p; 320 00m; 40V; 42, 15, 32, 9p; 320 000; 50V: 47 12p; 68 20p; 220	BC143 30 BD517 75 MPF102 66 T1590 30 213706/7 10 31128 112 BC147/8 9 BD695A 99 MPF103 36 T1591/93 32 213706/9 10 31140 112 BC147/8 10 BD695A 99 MPF103 36 T1591/93 32 213706/9 10 31140 112
9p; 47 8p; 100 11p; 150 12p; 220 15p; 330 22p; 470 25p; 680, 1000 44p; 2200 50p; 3300 76p; 4700 92p; 16V: 40, 47, 100 9p; 125 12p; 220 13p; 470 20p; 680 34p; 1000 27p; 1500 31p; 2200 36p; 3300 76p; 4700 79p; 1500 31p; 4700 70p; 4700 70p	BC148B 10 BDY56 180 MPF105 36 VK1010 80 2N371 10 40100 215 BC148C 10 BDY56 160 MPF106 40 VN10KM 45 2N3771 179 40101 130 BC149 9 BF115 35 MP5405 25 VN4664 78 2N3771 179 40101 130
TAG-END TYPE: 70V: 4700 245p; 64V: 3300 198p; 2200 139p; 50V: 3300 164p; 2200 110p; 40V: 4700 360e; 25W Log & Linear Values.	8C149C 12 8F167 29 MPSA06 25 VN66AF 85 2N3773 270 40251 97 8C153/4 27 8F173 27 MPSA12 32 VN88AF 85 2N373 270 40251 97 8C157/8 10 8F177 25 MPSA55 30 TTV107/8 11 20900 45 2013 130
TANTALUM BEAD CAPACITORS: 5000, 1K & 2K (LIN ONLY) Single 29p 550-2MD single gang 29p 550-2MD single gang 29p 550-2MD single gang 29p	BC159 11 8F178 30 MPSA56 30 ZTX109 12 ZN362 95 40315 60 8C1660 45 8F179 35 MPSA70 30 ZTX212 28 ZN3866 90 40316 95 8C1674 10 8F189 78 MPSA70 30 ZTX212 28 ZN3866 90 40316 95
18p; 4-7, 6-8 22p; 10 28p; 16V: 2-2, 3-3, 16p; 4-7, 6-8, 10 18p; 15 36p; 22 30p; 33, 47 40p; 100 75p; 220 88p 10V: 1W Wire-wound 50Ω-20K 115p 15, 22 26p; 33, 47 35p; 100 55p.	8C168C 10 8F194/5 12 MPSU05 55 ZTX303 25 2N3905 15 4036/2 40 8C169C 10 8F196/7 12 MPSU05 55 ZTX303 25 2N3905 15 4036/2 50 8C170 15 8F196/7 12 MPSU05 55 ZTX303 25 2N3905 17 40361/62 50
SILVER MICA (pF) SIEMENS multilayer miniature 0.25W log and linear values 60mm track 2, 3-3, 4-7, 6-8, 8-2, 10, 12, capacitors. 5KG-560KO Single gang 70p	BC177/2 11 8F199 18 MPSU55 60 ZTX314 25 244058 10 40407 60 BC173 11 8F200 30 MPSU55 60 ZTX320/26 30 2N4051/2 10 40407 60 BC177 10 40407 60 2020/2 10 40407 60 2020/2 10 40407 60 40407
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220, 250, 270, 330, 360, 39n, 56n, 68n 9p, 390, 470, 600, 800 & 100V: 100n, 120n, 10p; 150n 820nF 210 arch 11 220n, 130n, 130n, 10p; 150n 0.1W 500-2.2M Mini Vert. & Horiz, 7p 0.25W 1000-3.3MG Horiz, larger 10p	BC184/3 10 BF256B 35 0C41/42 120 ZTX504 25 2N4871 55 40468 85 BC184 10 BF257/8 32 0C44 120 ZTX531 25 2N5135/6 20 40594 105 BC182L 10 BF259 35 0C45/70 40 ZTX550 25 2N5138 18 40595 110
1000, 1200, 1800 30p each 470n 23p; 680n 30p; 1μ ² 3300, 4700 60p each 34p; 2μ2 50p. 1μ ² Precision Cermet 1W 1000-100K 80p	BC1841 10 BF274 42 CC71/72 40 2N696 30 2N5172 18 40603 110 BC1841 10 BF336 40 C76 50 2N697 23 2N5179 45 40636 175 BC187 26 BF451 35 CC81/82 50 2N698 40 2N5180 45 40673 95
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56nF, 100, 200n 9p 0-50V 21/2<-10/7 E/2 1p 50V: 470nf 12p 0-50V 20/2<-4/7 E/2	6C214 10 BFR41//9 23 IIP29C 36 2N914/5 32 2N5457 36 8C2141 10 BFR60/81 25 IIP29C 32 2N914/5 35 2N5458/9 36 8C2137/8 14 BFR98 105 TIP30A 48 2N930 20 2N5485 36
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80, 0.3W, 2": 2.25", 2.5", 3" BOD LEDS including Clips	VICS 0.5" LIQUID CRYSTAL DISPLAYS	SPECIAL OFFE	R DIL S	DIL SDCKETS (TEXAS) WATFORD'S			
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6A/8007 65 MVAM2 185 195p + 50p p&p Moise Diode BA102 30 BALOE ETCH RES Z5J 195 BB105 40 Pen + Spare tip	8x6x3" 210 10x41x3" 240 10x7x3" 275 12x5x3" 260 90p 12x8x3" 295	an edge over competitors.	OUI mini. Sold	CONNECTORS (Cannon type er Bucket Plugs Sockets	6-0-6V 100mA: 9-0-9V 75mA; 12-0-12V 75mA 6VA type: 6V-5A 6V-5A; 9V-4A 9V-4A; 6V-3A 12V-3A 12V-3A; 15V-25A 15V-25A 220p plastic 12VA; 45:13A 4-5V-13A; 6V-12; 6V-1 plastic 12VA; 4-5:13A 4-5V-13A; 6V-12;		
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TELEPHONE (STD 0277) 230909 international 200 North Service Road, Brentwood, Essex



10 MHZ Bandwidth P4 Standard £59.50+VAT (£68.42) plus carriage £3.00 For P31 (green) high resolution tube add £12.50+VAT (£14.38)





Statements and

-

24 MHZ Bandwidth P31(green)Standard for only £78.00 + VAT (£89.70) plus carriage £3.00

ALL MAJOR CREDIT CARDS ~ ACCEPTED - Small surcharge **CROFTON ELECTRONICS LTD** 35 Grosvenor Road, Twickenham, Middx TW1 4AD

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Now the time can tell you! .

from Silicon Speech Systems (a Powertran New subsidiary) - the first ever easy-to-build kit that will give a whole new meaning to the 'speaking clock'! Electronics and quartz technology combine to enable you to construct a talking timepiece that is interesting to build - fun to have!



RING ANDOVER (0264) 64455. Goods normally despatched 7 days

ELECTRONIC IGNITION K11



TOTAL ENERGY DISCHARGE electronic ignition gives all the well known advantages of the best capacitive

- discharge systems. PEAK PERFORMANCE ------ higher output voltage under all conditions.
- IMPROVED ECONOMY ----- no loss of ignition performance between services
- FIRES FOULED SPARK PLUGS no other system can better the capacitive discharge system's ability to fire fouled plugs.
- ACCURATE TIMING - prevents contact wear and arcing by reducing load to a few volts and a fraction of an amp.
- SMOOTH PERFORMANCE -- immune to contact bounce and similar effects which can cause loss of power and roughness.

PLUS

- 3½ times the energy of ordinary SUPER POWER SPARK capacitive systems - 31/2 times the power of inductive systems.
- **OPTIMUM SPARK DURATION 3 times the duration of ordinary** capacitive systems - essential for use on modern cars with weak fuel mixtures.
- BETTER STARTING ------ full spark power even with low battery
- CORRECT SPARK POLARITY unlike most ordinary C.D. systems the correct output polarity is maintained to avoid increased stress on the H.T. system and operate all voltage triggered tachometers.
- L.E.D. STATIC TIMING LIGHT for accurate setting of the engine's most important adjustment.
- LOW RADIO INTERFERENCE fully suppressed supply and absence of inverter 'spikes' on the output reduces interference to a minimal level.
- SIGNED IN RELIABILITY an inherently more reliable circuit combined with top quality components plus the 'ultimate insurance' of a changeover switch to revert instantly back to standard ignition. DESIGNED IN RELIABILITY

IN KIT FORM

it provides a top performance electronic ignition system at less than half the price of competing ready built systems. The kit includes everything needed, even a length of solder and a tiny tube of heatsink compound. Detailed easy-to-follow instructions, complete with circuit diagram, are provided - all you need is a small soldering iron and a few basic tools.

AS REVIEWED IN

ELECTRONICS TODAY INTERNATIONAL JUNE '81 ISSUE and EVERYDAY ELECTRONICS DECEMBER '81 ISSUE

FITS ALL NEGATIVE EARTH VEHICLES. 6 or 12 volt, with or without ballast

OPERATES ALL VOLTAGE IMPULSE TACHOMETERS Some older current impulse types (Smiths pre '74) require an adaptor -**PRICE £2 95**

STANDARD CAR KIT £14.85 ASSEMBLED AND TESTED £24.95

TWIN OUTPUT KIT £22.94 For MOTOR CYCLES and CARS with twin ignition systems

ASSEMBLED AND TESTED £34.70 Prices include V.A.T.

ECTRONIZE DESIGN Dept. B



Magnus Road, Wilnecote, Tamworth. B77 5BY Phone 0827-281000

BARCLAYCARD VISA

PLUS £1 U.K. P.&P. DIMENSIONS:

Length 12.5 cm Width 8.9 cm 4.3 cm Height Lead length 100.0 cm

TECHNICAL DETAILS

The basic function of a spark ignition system is often lost among claims for longer 'burn times' and other marketing fantasies. It is only necessary to consider that, even in a small engine, the burning fuel releases over 5000 times the energy of the spark, to realise that the spark is only a trigger for the combustion. Once the fuel is ignited the spark is insignificant and has no effect on the rate of combustion. The essential function of the spark is to start that combustion as quickly as possible and that requires a high power spark.

The traditional capacitive discharge system has this high power spark but, due to it's very short spark duration and consequential low spark energy, is incompatible with the weak air/fuel mixtures used in modern cars. Because of this most manufacturers have abandoned capacitive discharge in favour of the cheaper inductive system with it's low power but very long duration spark which guarantees that sooner or later the fuel will ignite. However, a spark lasting 2000µS at 2000 rev/min. spans 24 degrees and 'later' could mean the actual fuel ignition point is retarded by this amount.

The solution is a very high power, medium duration, spark generated by the TOTAL ENERGY DISCHARGE system. This gives ignition of the weakest mixtures with the minimum of timing delay and variation for a smooth efficient engine.

- SUPER POWER DISCHARGE CIRCUIT A brand new technique prevents energy being reflected back to the storage capacitor, giving 31/2 times the spark energy and 3 times the spark duration of ordinary C.D. systems, generating a spark powerful enough to cause rapid ignition of even the weakest fuel mixtures without the ignition delay associated with lower power 'long burn' inductive systems.
- HIGH EFFICIENCY INVERTER A high power, regulated inverter provides a 370 volt energy source - powerful enough to store twice the energy of other designs and regulated to provide sufficient output even with a battery down to 4 volts.
- PRECISION SPARK TIMING CIRCUIT This circuit removes all unwanted signals caused by contact volt drop, contact shuffle, contact bounce, and external transients which, in many designs, can cause timing errors or damaging un-timed sparks. Only at the correct and precise contact opening is a spark produced. Contact wear is almost eliminated by reducing the contact breaker current to a low level - just sufficient to keep the contacts clean,

TYPICAL SPECIFICATION

	ENERGY	DISCHARGE	
SPARK POWER (PEAK)	140 W	90 W	
SPARK ENERGY (STORED ENERGY)	36 mJ 135 mJ	10 mJ 65 mJ	
SPARK DURATION	500 µS	160 µS	
OUTPUT VOLTAGE (LOAD 50pF			
EQUIVALENT TO CLEAN PLUGS)	38 KV	26 K V	
OUTPUT VOLTAGE (LOAD 50pF + 500 KC	2		
EQUIVALENT TO DIRTY PLUGS)	26 KV	17 KV	
VOLTAGE RISE TIME TO 20 KV			
(Load 50pF)	25 µS	30 µS	
TOTAL ENERGY DISCHARGE should not	be confused	with low pow	1

TOTAL ORDINARY

inductive systems or hybrid so called reactive systems. Goods normally despatched within 7 days.

PRACTICAL ELECTRONICS - STEREO This easy to build 3 band stereo AM/FM tuner kit is designed in conjunction with Practical Electronics (July issue). For ease of construction and alignment it incorporates three Mullard modules and an I.C. IF. System,

FEATURES: VHF, MW, LW Bands, interstation muting and AFC on VHF. Tuning meter. Two back printed PCB's. Ready made chassis and scale. Aerial: AM - ferrite rod, FM - 75 or 300 ohms. Stabilised power supply with 'C' core mains transformer. All components supplied are to P.E. strict specification. Front scale size 101/2" x 21/2" approx. Complete with diagrams and instructions.

SPECIAL OFFER! TUNER KIT PLUS

Matching I.C. 10+10 Stereo Po amplifier kit (usually £3.95 + £1.15 p&p) Mullard LP1183 built preamp, suitable for magnetic/ceramic and auxiliary inputs (usually £1.95 + 70p p&p) • Matching power supply kit with trans-former (usually £3.00 + £1.95 p&p)

Matching set of 4 slider controls complete with knobs for bass, treble and volumes (usually £1.70 + 80p p&p)

F

21.9

ATVE

STEREO AMPLI

Featuring latest SGS/ATES TDA 2006 10 watt output IC's with in-built thermal and short circuit protection.
 Mullard Stereo Preamplifier Module.

Attractive black vinyl finish cabinet, 9"x 8%"x 3%" (approx)
 10+10 Stereo converts to a 20 watt Disco amplifier.

To complete you just supply connecting wire and solder. Features include din input sockets for ceramic cartridge, micreactors include on input sockets for certainty cartrogy, inc rophone, tape or tuner. Outputs - tape, speakers and head-phones. By the press of a button it transforms into a 20 watt mono disco amplifier with twin deck mixing. The kit incorp-orates a Mullard LP1183 pre-amp module, plus power amp assembly kit and mains power supply. Also features 4 slider level controls, rotary bass and treble controls and 6 push button switches. Silver finish fascia with matching knobs and

contrasting cabinet. Instructions available, price 50p. Supplied FREE with the kit.

£14.95 Plus £2.90 p&p SPECIFICATIONS: Suitable for 4 to 8 ohm speakers.

requency response Input sensitivity

40Hz - 20KHz. P.U. 150mV. Aux. 200mV. Mic. 1.5m V. Bass ±12db @ 60Hz Treble ±12db @ 10KHz 0.1% typically @ 8 watts 220 - 250 volts 50Hz.

Tone controls Distortion Mains supply

STEREO MAGNETIC PRE-AMP CONVERSION KIT Includes FREE Magnetic cartridge with diamond styli. All components including p.c.b. to convert your ceramic in-put on the 10+10 to magnetic. Only available with 10+10 amp. £2.00 includes p&p.

8" SPEAKER KIT Two 8" twin cone domestic speakers, £4.75 per stereo pair plus £1.70 p&p. when purchased with amplifier. Available separately £6.75 plus £1.70 p&p.

PRACTICAL ELECTRONICS SERIES II CAR RADIO KIT

2 WAVE BAND MW -- LW

· Easy to build + 5 push button

tuning . Modern design . 6 watt output . Ready etched

and punched PCB . Incorporates suppression circuits All the electronic components to build the radio, you supply only the wire and the solder, featured in Practical Electronics March issue. Features: pre-set funing with 5 push button options, black illuminated tuning scale. The P.E. Traveller has a 6 watt output neg, ground and incorporates an integrated circuit output stage, a Mullard IF Module LP1181 ceramic filter type pre-aligned and assembled, and a Bird pre-aligned push button tuning unit. Plus f2.00 p&p.

Suitable stainless steel fully retractable aerial (locking) and speaker (6"x 4"app.) £1.95/pack. Plus £1.15 p&p available as a kit complete



IGH POWER READY BUILT OR IN KIT FORM

plus £3.80 p&p.

ONEY

BUILT

£18.95

£14.25 Plus £1.15 p&p

Plus £1,15 p&p

200 W Model

200 watts 70 - 95 max

4 · 16 ohms

25Hz - 20KHz

400mV @ 47K

7.95

KIT £10.50 125 WATT MODEL Plus £1.15 p&p 200 WATT MODEL £14.95 Plus £1.15 p&p SPECIFICATIONS: 125 W Model

Max, output power (RMS) Operating voltage (DC) Loads Frequency response

measured @ 100 watts 25Hz - 20KHz Sensitivity for 100 watts 400m V @ 47K Typical T.H.D. @

50 watts, 4 ohms 0.1% 0.1% Dimensions (both models) 205 x 90 and 190 x 36mm

The power amp kit is a module for high power applications — disco units, guitar amplifiers, public address systems and even high power domestic systems. The unit is protected against short circuiting of the load and is safe in an open circuit condition. A large safety margin exists by use of

125 watts

Δ - 16 ohms

50 + 80 max



30+30 WATT STEREO AMPLIFIER

Viscount IV unit in teak simulate cabinet, silver finished rotary controls and pushbuttons with matching fascia. mains indicator and stereo jack socket. Functions with for mic magnetic and crystal pickups, tape and auxiliary. Rear panel features fuse holder, DIN speaker and input socket 30+30 watts RMS, 60+60 watts peak. For use with 4 to 8 ohm speakers £32.90 Size 14%"x10" approx BUILT AND TESTED

Plus £3.80 p&p

TV SOUND TUNER KIT as featured in E.T.I. December '81 issue. Kit of parts including PCB, UHF tuner, I.C.'s, all components excluding case, and selector switch. £11.45 + £1.50 p&p.



Transformer £1.50 + £1.50 p&p (p&p free on trans former if ordered with kit). • Ready built LP1183 Module for simulated stereo operation £1.95 + 75p p&p.

CALLERS ONLY

323 Edgware Rd, London W2. Tel: 01-723 8432. Open 9.30am - 5.30pm. Closed all day Thursday. Persons under 16 not served without parents authorisation

ALL PRICES INCLUDE VAT AT 15%

All items subject to availability. Prices correct at 1/12/81 and subject to change without notice. RTVC Limited reserve the right to update their products without notice.

generously rated components, result, a high powered rugged unit. The PC Board is back printed, etched and ready to drill for ease of construction and the aluminium chassis is preformed and ready to use. Supplied with all parts, circuit diagrams and instructions.

plus £2.50 p&p

P. E. QUAL

ACCESSORIES:

Suitable LS coupling electrolytic for 125W model	£1.00 plus 25p p&p.
Suitable LS coupling electrolytic for 200W model	£1.25 plus 25p p&p.
Suitable mains power supply unit for 125W model	£7.50 plus £3.15 p&p.
Suitable Twin transformer power supply for 200W model	£13.95 plus £4.00 p&p

MONO MIXER AMPLIFIERS



50 WATT six individually mixed loputs for two pick ups DU WAT I Six individually mixed inputs for two pick ups (Cer, or Mag,), two moving coil microphones and two aux-iliary for tape, tuner, organs, etc. Eight slider controls - six for level and two for master bass and treble, four extra treble controls for mic and aux inputs. Size: 13% "x5%" x3%" (app. Power output 50 watts R.M.S. (continuous) for use with 4 to

8 ohm speakers. Attractive black vinyl case with matching fascia and knobs. Ready to use

£39.95 Plus £3.70 p&p



Bruched Aluminium fascia and rot-

100 WATT

ary controls. Size: approx. 14"x4"x10%". Five vertical slider controls, master volume, tape level, mic level, deck level, PLUS INTERDECK FADER for perfect graduated change from record deck No. 1 to No. 2, or vice versa. Pre fade level controls (PFL) lets YOU hear the next

c before fading it in. £76.00 VU meter monitors output 100w RMS output (200w peak). Plus £4.60 p&p

MAIL ORDER ONLY

218 HIGH STREET, ACTON, W3 6NG. Note: Goods despatched to UK postal addresses only. For further information send for instructions 20p plus stamped addressed envelope. All goods delivered within 14 days of receipt of order.



FREE

While reading the local paper recently, the thought occurred that there must be hundreds of electronics hobbyists with odd bits of test gear, audio equipment, or components stacked away that they could sell in an "under a tenner" column of the type now found in many local papers. During discussions in the editorial office it was thought to be an excellent idea and that we could run advertisements with no price limit, provided the service was only available to private readers and not to the trade or anyone making money from buying and selling anything to do with electronics. After all, the trade can advertise in our pages or take a classified space, but with the cost of classifieds rising it is not a viable proposition for the reader who would like to sell a few components, a speaker or a meter.

We think a readers free ad. page will be of interest to most readers, will encourage an interface between hobbyists and could realise some cash to help with our hobby. We want to encourage you to buy, sell or swap anything to do with electronics from a bag of resistors to a printer, computer, organ, TV or oscilloscope but we have decided that we cannot accept ads. for computer software. After all, if you're writing and selling software you are "in business" and if you're selling someone else's you're probably infringing their copyright!

PE BAZAAR

We have decided to call the feature *PE Bazaar* as we feel this term describes perfectly just what it's all about. As we have said it is for **PE readers only** and we have thus instigated a coupon system for sending in ads. This means that you must have an up to date issue to send in your ad. and you must sign a declaration to say you are not running a business dealing with electronics in any way. We have reserved the right to refuse ads. if we are worried about the contents or the private reader/business situation.

In order to give everyone a chance we can only accept one ad. per coupon and .we have limited the number of words in each ad. so that we can get a good variety on a page, or in any small space we may have—say at the end of an article. Full details of *PE Bazaar* and the first ad. coupon can be found on page 31. If we get some ads. back very soon after the publication of this issue, we hope to be able to get them in next month—it's first come, first served.

RESPONSIBILTY

One final point, we must make it quite clear that PE cannot be held responsible for any errors in the free ads. or for any transactions that take place between readers as a result of a free ad. We will not enter into correspondence concerning free advertisements, their contents or transactions.

However, we must say that in our editorial dealings with our readers we have found you to be a wonderfully reliable, trustworthy and honest lot. We have often sent odd items to readers, requesting payment on receipt and almost without exception have received prompt payment. We would like to take this opportunity to thank you—it saves us all time, trouble and money.-when things work in this way. Let's hope *PE Bazaar* is popular, helps as many of you as possible and leads to some useful exchanges and friendships within our hobby. It's up to you now!

Like everything else, it seems to be such a good idea, we wonder why we didn't think of it months ago?

Mike Kenward

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We regret that lengthy technical enquiries cannot be answered over the telephone (see below).

Advertising Offices: Practical Electronics Advertisements, King's Reach Tower, King's Reach, Stamford Street, SE1 9LS Telex: 915748 MAGDIV-G

Technical Queries

We are unable to offer any advice on the use or purchase of commercial equipment or the incorporation or modification of designs published in PE. All letters requiring a reply should be accompanied by a stamped, self addressed envelope, or international reply coupons, and each letter should relate to one published project only.

Components and p.c.b.s are usually available from advertisers; where we anticipate difficulties a source will be suggested.

Back Numbers

Copies of most of our recent issues are available from: Post Sales Department (Practical Electronics), IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 OPF, at 95p each including Inland/Overseas p&p.

Binders

Binders for PE are available from the same address as back numbers at £4.60 each to UK or overseas addresses, including postage and packing, and VAT where appropriate. Orders should state the year and volume required.

Subscriptions

Copies of PE are available by post, inland or overseas, for £13.00 per 12 issues, from: Practical Electronics, Subscription Department, Oakfield House, Perrymount Road, Haywards Heath, West Sussex RH16 3DH. Cheques and postal orders should be made payable to IPC Magazines Limited.

B REVEN

Jasper Scott

Public consultation on radio frequencies

Wider public consultation on proposals which could affect frequency allocations for some aircraft and shipping radio services has been announced by the Home Office.

The Home Secretary was giving details of the United Kingdom's preparations for the 1983 World Administrative Conference for Mobile Telecommunications, to be held in Geneva, in answer to a Parliamentary Question from Mr. Peter Viggers MP. The decisions made at the Geneva Conference will govern frequency allocations for certain maritime and aeronautical mobile radio services.

The Home Secretary pointed out that nongovernment organisations had already been consulted about the UK's provisional proposals for the conference, but he had decided that views should be sought from a broader public so that no significant point was overlooked. The UK proposals deal mainly with two major items:

Edited by

Regulatory provisions for the introduction and development of the Inter-Governmental Maritime Consultative Organisation's Future Global Distress and Safety System; and the drawing-up of channelling plans for the maritime mobile service in the mediumfrequency (MF) and high-frequency (HF) bands. Other UK proposals seek improvements in the regulation of the aeronautical mobile service, and cover miscellaneous points on the agenda.

Copies of the proposals can be obtained from: Home Office Radio Regulatory Department, Room 804, Waterloo Bridge House, Waterloo Road, London SE1 8UA.

Grand Prize

British Telecom's Prestel world viewdata service is offering a £1000 prize to the designer of the best Prestel adaptor for the top-selling Sinclair ZX81 personal computer.

Sinclair Research, which is to provide one of the judges, strongly supports the competition and believes that a successful design will offer its established 150,000 UK users a valuable new application. Although a number of microcomputers can already receive Prestel, these numbers would mark a very significant service extension for British Telecom.

Telesoftware—computer programs distributed via teletext or viewdata to computers in schools, homes and offices—is now a major Prestel growth area, and many program publishers are becoming active and establishing libraries that can hold up to 1,000 pages following the Department of Industry's initiative in funding the Educational Telesoftware Project.

The prize will be awarded to the adaptor which best combines 'low price, elegant design and practical robustness', and the working design submitted must be capable of being modified to receive approval for attachment to the telephone network.

Closing date is March 14th, 1982, and further details, specification and entry forms are available from Tony Sweet, Prestel Headquarters, Telephone House, Temple Avenue, London EC4 0HL (01-583 9811).

AM CB: AN APRIL FOOL

Following numerous rumours that the Government is considering legalising CB on 27MHz AM, the Home Office has issued the following statement:

"Don't be misled by unfounded rumours claiming that the use of illicit 27MHz AM sets will be legalised. The Government has no intention of making any changes to the new legal 27MHz FM CB service."

Reports abound of AM sets carrying labels stating that the apparatus cannot be used "until April 1982". Any such stickers are quite simply hoaxes.



Can't quite make out that dry joint or broken p.c.b. track? What you may need is a pocket microscope, and Stotron Ltd. have recently introduced one which could be just what you need. With 20X magnification and a graticule showing linear and angular measurements, the unit is only 125mm long. Illumination is powered by standard 1.5V penlight batteries, and a micro-stand with spring clips for sample slides is also available, so the microscope can be used like a conventional model.

Priced at £16.99 plus £2.80 for the stand, the 'scope is available from Stotron Ltd., Unit 1, Haywood Way, Ivyhouse Lane, Hastings, East Sussex (0424 442160).

CAR COMPUTER

Pimac Systems Ltd., who supply kits for the PE Car Computer, inform us that they have been able to reduce the price for the complete kit to £78.50 plus £1 p&p. The unit is also available ready built, price £88.50 plus £1 p&p. Both prices include VAT. Pimac Systems Ltd., 20 Bloomfield Road, Moseley, Birmingham B13 9BY.



Items mentioned are available through normal retail outlets unless otherwise specified. Prices correct at time of going to press.

Briefly....

This year, the All-Electronics/ECIF show will be doubled in size and held at the Barbican Exhibition Centre in the City of London. It will run from Tuesday April 20th to Thursday April 22nd, and, say the organisers. will offer the largest and most comprehensive display of the electronics industry that the capital has seen for many years.

. . .

Tempus—often featured in these pages as suppliers of Casio products—have moved from their premises in East Road, Cambridge. Their new address is: 38 Burleigh Street, Cambridge CB1 1DG.

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Barrie Electronics Ltd. inform us that they have been appointed a franchised distributor for the "budget range" of toroidal transformers manufactured by Cotswold Electronics Ltd.

The fifty-eight toroidal transformers will complement the 150 stacked laminated types stocked by Barrie Electronics on an off-theshelf basis.

Barrie Electronics Ltd., 3 The Minories, London EC3N 1BJ (01-488 3316).



Due to demand for this exceptionally priced tool which was on Special Offer in our January issue, the offer closing date has been extended to March 26th 1982. Please contact Watford Electronics to ensure supply. Price is £8.60 inclusive.

WERSI'S COMET



Recently unveiled by Aura Sounds in London, the Wersi Comet is the latest addition to the increasingly popular range of Wersi Organs. The Comet is a completely new model—not a modification or development of any other Wersi organ, and has been designed to give the best possible sound quality and features at a budget price.

As well as traditional drawbar and orchestral sounds, the Comet provides a wide range of other voices, including various guitars, synthesiser and percussion. Numerous effects are incorporated, including three way vibrato, slalom and the Wersivoice rotating speaker sound. A totally new feature is the Comet's facility to accept connection of up to four 'satellite' keyboards, so that up to five musicians can play the one organ together.

As with all other Wersi instruments, the Comet is available ready built or as a kit. If you choose to build a Comet yourself (which is a very straightforward operation estimated to take only 100 hours, due to the use of plug-in circuit boards and ribbon cable wiring) you can save almost half the cost of the ready built unit.

The Comet is available either as a home console model (W10S) or as a portable unit with detachable legs (W10T). Prices are as follows: W10S £3,620 ready built; £1,899—complete kit. W10T £3,592—ready built: £1,971 complete kit. Further information is available from Aura Sounds Ltd., 14–15 Royal Oak Centre, Brighton Road, Purley, Surrey (01-668 9733).

Goundoun...

Please check dates before setting out, as we cannot guarantee the accuracy of the information presented below.

BEX Bournemouth Feb. 17-18. K

Microsystems Feb. 24-26. West Centre Hotel, London. Z1

Seminex Mar. 29-Apr. 2. Imperial College, London. H1

Laboratory Edinburgh Mar. 30-31. Ass. Rooms, Edinburgh. E

CAD Mar. 30-Apr. 1. Metropole, Brighton. Z1

Sensors & Systems Mar. 30-Apr. 1. The Forum, Wythenshawe, Manchester. T

ETM Mar. 30-Apr. 1. The Forum, Wythenshawe. T

Peripherals Mar. 31-Apr. 2. West Centre Hotel, London. Z1

Laboratory Manchester Apr. 7–8. New Century Hall, Manchester. E All Electronics Show Apr. 19–21. Barbican Centre, London. E BEX Brighton Apr. 28–29. K Compec Europe May 4–6. Centre Int. Rogier, Brussels. Z1 Scotelex. Jun. 8–10. Roy. Highland Ex. Hall, Ingliston, Edinburgh. A1 BEX Leeds Jun. 9–10. K Transducer/Tempcon Jun. 29–Jul. 1. Wemb. Conf. Centre. T BEX Croydon Jun. 30–Jul. K Leeds Electronics Show Jul. 6–8. University. E Laboratory London Sep. 14–16. Grosvenor House. E

- A1 Institute of Electronics, Rochdale, Lancs.
- E Evan Steadman, Saffron Walden & 0799 22612
- HI Seminex Ltd., Tunbridge Wells (0892 39664
- K Douglas Temple, Bournemouth & 0202 20533
- Li World Trade Centre & 01-488 2400
- T Trident Tavistock & 0822 4671
- V SDL & Dublin 763871
- Z1 IPC Exhibitions, Sutton & 01-643 8040



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Salaries

A new year always brings thoughts of what lies ahead and reflection on the past. A main preoccupation is the personal one of jobs and salaries. On the broader front there is the anxiety that political hot-heads will rock the industrial boat and put all our jobs at risk. But assuming, hopefully, that militants can be bought off or otherwise contained then 1982 could still be a good year for recovery.

Barring political upheaval, any competent person in electronics can expect to hold his job. The adventurous may well rapidly improve their position and prospects by changing jobs. There are plenty of openings available and industry personnel managers are still complaining that good people are hard to find and how much it costs in advertising and interview time to find them.

The lowest advertised salary I have spotted recently is for R & D engineers at £5,176 for a junior starting post. But this was in Government service carrying some advantage in security and rising to £8,589.

A BBC call for transmitter and planning engineers gave a range of £9,000 to £11,000. An instrument engineer in South Wales is offered circa £9,000, another similar post in Essex up to £12,000. A development engineer in television was promised £10,000, his laboratory supervisor £16,000.

Overseas the salaries are higher and most enjoy tax advantages. A quick sample threw up a salesman (Middle East) £18,000, telecomms engineer (Saudi Arabia) £20,000 tax free, electronics lecturers (Hong Kong) circa £20,000, branch manager (Iraq) up to £24,000.

Everyone will hold his own views on the value of the salaries in relation to environmental factors, separation from family etc. I have known people move job to a lower salary in a different area to secure better education for their children. It seems that salaries in general are holding up well and that the variety of jobs and locations in electronics gives far more choice and opportunity than most other industries.

Backward Glance

Last year was pleasurable in the number of success stories I was able to record. And it finished up with the giants of the industry continuing to improve not only their sales (including exports) and profits but also their order books. GEC-Marconi, for example, managed another scoop with a £500 million contract for the Royal Navy's heavyweight torpedo. On the technological front there were also gains with many new equipments and systems on offer.

Personality of the year was unquestionably Sir Ernest Harrison, mastermind of the Racal Electronics Group, who was heaped with honours, principally a well-deserved Knighthood but also Honorary Doctorates from Cranfield Institute of Technology and the University of Surrey. He finished the year with a flourish, winning the Businessman of the Year Award.

The citation of the latter mentioned the acquisition and subsequent tum-round in fortune of Decca but I think the last part would have pleased Sir Ernest best. It added, "And for providing a rallying point for industry, a much-needed demonstration that where the will is present, British companies, their managements and their products can equal the best in the world, even in the toughest high-technology markets which are too often considered to be the preserve of Japan and the USA."

Another well-deserved honour, the Gold Medal of the Royal Aeronautical Society, went to J.E. (Jack) Pateman for 'building-up British technical and industry leadership in the avionics field'. Pateman is managing director of Marconi Avionics and in ten years he created 8,000 new jobs, tripling the size of the company to become Europe's largest producer and top exporter of aviation electronics.

In case you imagine these men are nothing but workaholics I should mention that Jack Pateman enjoys gardening and sailing and Sir Ernest shares Jack's enthusiasm for gardening but also lists wild-life, all sports, but particularly soccer (I recall he was an Arsenal fan at one time and maybe still) among his interests.

Up North

With the great bulk of the electronics industry concentrated in the South and South East, the Midlands and Northern regions get less attention than they deserve. I recently had the opportunity of visiting three companies in regions which are generally described as depressed. This is certainly true of the inner city areas I saw on the trip but there was nothing depressing in the companies.

Maybe I was lucky in my choice but all three had expanded during the recession, had new products and services in planning, and all believed in a great future. One of the companies, to my astonishment, had been running three shifts round the clock throughout the whole of the recession. None of them had had an easy time but there were no complaints. Just enthusiasm to get on with the job.

Energy

It seems only yesterday that we were all talking of the energy famine. Today we have a glut of oil and coal. But yesterday's scare triggered a surge of activity in developing alternative sources of energy.

One result is IBA's experimental sunand-wind powered TV relay station at Bossiney on the north Cornish coastline. All power for the transposers and receivers come from natural sources. The idea of a hybrid sun-and-wind system is that if the sun isn't shining there is more than a fair chance that the wind will be blowing, and often both will be generating at the same time. When nature is not generous the system is backed up by a reservoir of three dozen lead-acid batteries with a capacity of 1,000Ah. The windmill produces 150W at wind speeds of 15 m.p.h. and 864 solar cells produce up to 780W in peak sunlight. The TV relay equipment consumes 150W.

To get Channel Four available to a target of 99.9 percent of the population some extra hundreds of such low power relay stations will be needed. But energy conservation is only part of the reason for the selfcontained station. The other, often more compelling, reason is that these low-power relays, sometimes serving as few as 500 people or less, are on difficult and remote sites where the capital cost of running power lines to them is out of proportion.

Inmos

The Inmos factory at Newport, Gwent, some 85,000 sq.ft. of it, is now becoming a reality. Building has been in progress for a full year and the 30,000 sq.ft. clean area should be installed by March.

The Newport plant appears to be definitely scheduled for wafer fabrication and an eventual workforce of 1,000 is still envisaged. But assembly might still take place in the Far East.

A number of questions still overshadow the Inmos project, not least the departure of Paul Schroedor who as Executive Director, corporate development, was clearly a key man. The product strategy and future funding remain unclear. Obviously there are elements of commercial security involved. Anyway, with so much now invested and a tentative foothold in the market we can do no more than wish Inmos the best of British luck.

Vigorous ATE

Automatic Test Equipment (ATE) is still a blossoming market under vigorous attack by all the leading manufacturers. Typical is Marconi Instruments System 80X, unveiled at the ATE Exhibition in Brighton last December. The 80X has been designed with the US market in mind and MI has opened an ATE facility at Sunnyvale in an attempt to gain a larger market share. Prices start at £80,000 and rise to £150,000 for a complex system with up to 2,000 test points. THE cost of electrical energy consumption has become increasingly significant in recent years as the price of primary fuels has risen more quickly than the average increase in the cost of living. In order to control costs in the house, it is necessary to know which appliances cost the most to use and to what extent savings can be made by reduced usage, turning thermostats down, etc. The watthour meter provided by the electricity supply authority does give a means of calculating total consumption and hence the staggering quarterly bill. However, it gives no indication of the rate at which costs are being incurred or which appliances are consuming energy at the greatest rate and what the individual cost is.

All appliances are supplied with a label indicating their power rating but to calculate the cost of running a washing machine through a 45 minute cycle with intermittent motor running, thermostatically controlled heating, pumping and spinning would be almost impossible for the average person.

The device described in this article continuously measures the current being consumed by such an appliance and having previously had the latest cost per unit of electricity keyed into it, displays the charge as it is being incurred. It was designed about three years ago and is the subject of a UK Patent application. Only recently has the cost of the components used reduced to the level where a home construction kit has become an attractive proposition.

TELECTRIC BASICS

The basic operation of the Telectric can be explained with reference to the block diagram of Fig. 1. The current consumed by the appliance plugged into the 13A socket flows through the primary turns of the current transformer and the secondary voltage induced forms the input to IC3 in order to measure this current. The PSU provides a 5 volt supply and also a sync pulse which occurs at the zero crossing of the a.c. mains voltage. The microprocessor reads the control keys, provides the display drive and multiplexing and samples the output of IC3 at the peak of the mains voltage waveform on both negative and positive half cycles. From the stored value of the cost per unit the incremental cost per half cycle is calculated and added to the running total for display.

There is one control switch and four push button keys. When the switch is in the cost per unit position the display initially reads Ent. and a single push of any key increments the display digit directly above it in order to enter the cost per unit in pence to three decimal places. When the control switch is in the electricity bill position the four push buttons assume their control functions. These are:

Start Resets the accumulated charge and elapsed time to zero and starts new sequence.

Pause Holds the current reading of charge and time. A second push of this key restarts from the previous readings.

Time/Cost Toggles the display between displaying time and charge.

Check Whilst held down, the current value of the cost per unit is displayed as a check on operation.

DISPLAY FORMAT

In the cost per unit position the decimal point is located between the two left hand digits and the display indicates the unit cost in pence.

With the control switch in the electricity bill position the display is auto ranging, the device having been designed to accumulate up to £99.9 in cost and 99.9 days in time. The ranges and examples of their display formats are as follows: **Cost**

up to 9.99p	e.g. 2.14P	pence and decimals of
10p to 99.9p	e.g. 47.1P	pence
up to 9.99p	e.g. 2.14P	pence and decimals of
1 Op to 99.9p	e.g. 47.1P	pence
£ 1.0 to £9.99	e.g. L7.85	pounds and pence
£ 10.0 to £99.9	e.g. L90.6	pounds and decimals of
	0	pounds

Time

minutes and seconds
hours and minutes
days and decimals of days

CIRCUIT DESCRIPTION

All the circuitry in the device uses a 5 volt supply and the circuit diagram is divided into two parts; Fig. 2 showing the power supply and sync pulse generator and Fig. 3 showing the processor, etc. The 5 volts is generated using a 9V (6VA) transformer, bridge rectifier and a 7805 i.c. regulator. The sync pulse generation can be described with reference to Fig. 5 which shows the waveforms through IC8. The full wave rectified input to IC8a is inverted and produces a positive going pulse about 2ms wide and symmetrical about the zero crossing of the a.c. mains waveform. This is delayed by R21, C6. The output of IC8b is a negative going 2ms pulse with its leading edge corresponding approximately

etricity COST MONITOR

B.Sc. C.Eng. M.I.E.E.

SPECIFICATION

Display

4 digit 7 segments I.e.d.s 0.5" high with decimal points.

Controls

S1 Cost per Unit/Electricity Bill Output socket power switch (No effect on metering function)

S2-S5

Functions

A) Switch in 'Cost per Unit' position S2 Most significant digit increments every time

key is pressed, 0–9
 S3, 4, 5
 As S2 for digits 2, 3, 4 of pence per unit figure. This position of the switch serves only to enter the pence per unit figure. The decimal point second from left is

illuminated. The electricity cost is not monitored. Cost and elapsed time functions are set to zero

B) Switch in 'Electricity Bill' position S2 ('Check') While pressed pence per unit is displayed

	(operational check)
	Otherwise S5 controls display
S3 ('Start')	Each depression of key zeroes electricity bill
	and time elapsed. Electricity bill begins to
	total from moment of depression
	Subsequent depression zeroes and restarts.
S4 ('Pause')	Depression of key halts totalling of cost and elapsed time
	Subsequent depression restarts from previous reading without zeroing
S5 ('Time/	Display shows electricity bill initially. First
Cost')	depression causes display to show elapsed time. Second depression electricity bill, etc.

Maximum cost per unit 9.999 pence Maximum electricity bill total £99.9 Maximum elapsed time 99.9 days Accuracy better than 1% for 13A load

with the zero crossing. IC8c speeds up these edges and C8 couples them to IC8d which is biased below its threshold by R22 and R23. Thus the output from IC8d is a negative pulse approximately 20μ s wide corresponding to the zero crossing and this forms a synchronising pulse.

The microprocessor controlling the Telectric is the 8035 which contains its own clock generator, reset, and interrupt input pins. It has 64 bytes of on chip RAM, an 8 bit data/address bus, two 8 bit I/O ports and external memory read and write timing signals. It also has an on chip timer which is used to time the ADC sampling at the peak of the a.c. voltage waveform.





Details of the microprocessor, bus timing and software will follow next month, however, it is worth mentioning that the displays are of the common anode l.e.d. type and their pin connections are shown in Fig. 4. *Pnp* transistors are used to drive the common anode of each display and the multiplex frequency is 200Hz. The cathodes of the individual segments are driven via open collector gates and the microprocessor controls the display decoding in order to provide the special characters not normally available in a standard decoder.

CURRENT MEASUREMENT

In measuring mains current, the simplest way would be to use a series resistor but this has several drawbacks. The ADC requires a 5 volt peak-to-peak voltage waveform at its input for maximum current corresponding to a 40 amp peak to peak sinusoidal current or 14.14 amps r.m.s. The power developed in such a resistor would be 25 watts at peak load which is obviously not desirable. A smaller resistor could be used with an op-amp to multiply up the voltage generated but this technique still requires the circuitry to be connected to the mains and faults could be catastrophic. It was therefore decided to use a current transformer to provide full isolation.

Purpose designed current transformers are quite expensive and so a 3VA mains to low voltage transformer is used in the Telectric. This has its low voltage winding removed to enable one and a half turns of the mains neutral lead to be threaded onto the transformer before connection to the output socket.



Components . . .

Resistors

D1 010 010 014 016	101/16 off)
n1, n10, n12, n14, n10,	TOK (0 011)
R18	
R2 to R9	220 (8 off)
R11, R13, R15, R17	2.2k (4 off)
R19	47k
R20	470k
R21	1M
R22	100k
R23	220k
R2'4/R25	(see construction notes)
	٤.

Capacitors

		680
	C2, C3	22p (2 off)
	C4	1µ tant bead
	C5	0-1µ
	C6	680p
	C7	680μ 16V
	C8	470p
	C9	47µ tant bead
3 off 0.1µF disc ceramic and 2 off 0.01µF disc ceramic		
decoupling capacitors as per layout diagram.		

Semiconductors

Bridge rectifier (REC1)	50V, 1A
D1, D2	1N914 or equivalent
TR1 to TR4	BC213L or equivalent (4 of

Integrated Circuits IC1 2716 2K × 8 UVEPROM 1C2 74LS373 **IC3** ADC0804 (National Semiconductors) 8035 microprocessor **IC4** 1C5, 1C6 7401 (2 off) 74LS365 IC7 **IC8** 4069B 109 7805 (regulator + 5 volts) **Displays** DL507B (Litronix) Common anode 7 segment I.e.d. displays (4 off)

Miscellaneous

4 push buttons; 1 rocker switch; 1 switched 13A mains socket; 1 grey moulded ABS box, 190 × 110 × 60mm; 1 mains transformer 9 volt, 6VA, 10% regulation; 1 mains transformer 3VA to convert to current transformer; 1 crystal 3-579MHz (NTSC subcarrier); 1 heatsink for 7805; 2 metres 13A mains cable; 3 metres connecting wire; 8 i.c. sockets (3 × 14 pin, 1 × 16 pin, 2 × 20 pin, 1 × 24 pin, 1 × 40 pin); 1 p.c.b., double sided but designed for pin through as well as plated through holes; 1 display p.c.b.

A complete constructor's kit for the Telectric is available from Response Co., Stoner End, Froxfield, Petersfield, Hants (0730 3063). Price £47.50 + VAT (£54.62). A limited number of fully assembled Telectrics are also available at £64 + VAT. All parts are also available separately. Send SAE for details.



Fig. 4. The pin connections to a display digit



1 2 3

EG7 8 7

FUNCTION E CATHODE O CATHODE COMMON ANODE C CATHODE DP CATHODE B CATHODE A CATHODE G CATHODE G CATHODE

Pin

10



Internal assembly





CONSTRUCTION

All of the components except the display and switches are mounted on the main p.c.b. which is double sided. The foil layouts are shown in Figs. 6 and 7 and the component layout in Fig. 8. Although the p.c.b. is double sided it has been designed with no through connections on i.c. pins and the position of these is shown by large dots in Fig. 8. Also shown are some extra decoupling capacitors which do not appear as components in the circuit diagrams. The p.c.b. should preferably be assembled using i.c. sockets for all i.c.s so that faults can be easily rectified.



EP799

Fig. 6. Topside of p.c.b.

25



Fig. 7. Underside of p.c.b.



Fig. 8. Assembly details of board

The pads for the mains transformer have been made quite large to accommodate a variety of pin arrangements. The current transformer load resistors must be fitted before the ADC is plugged into the circuit. Without these resistors a high voltage may be generated which could damage or destroy the ADC. The values of R24 and R25 are found experimentally (they are supplied correct value with the kits) by using a one kilowatt load such as an electric fire. The exact current is measured with an Avo or similar and will be around 4 amps r.m.s. The voltage across the load resistors is then given by the relationship V(r.m.s.) = I $\times 0.125$ volts, where I is the exact current. R24 can now be selected to produce a slightly high voltage reading and R25 will be a higher value resistor to provide the final trim.

(R24 may be 390 and R25 10k giving a total equivalent load of 376).

NEXT MONTH Detailed discussion of the microprocessor circuitry and software and more construction details.

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T IS likely that most householders will experience at least one power failure. This can be very inconvenient, because it is then that the candles and matches can't be found, because of the dark! However, such a failure can be more serious, as the author found when carrying a tray of crockery on the stairs at the moment of failure. The shock alone was enough to send both flying, resulting in a painful fall. The idea for this project was thus conceived during convalescence. It obviously needed to be automatic, pleasing and cheap.

The project described here is intended to enable constructors to build a ni-cad powered emergency light, drawing no battery current in its quiescent state (less than 1mA from the rectified supply), switching on automatically in the event of a mains failure. When mains is restored, the ni-cad can then be recharged, and a built-in timer lights an I.e.d. when the batteries should be charged (assuming they were fully exhausted to start with). Two switches are provided, one to act as an on/off control and as a reset switch, the other to activate the charging and timing circuits.

HOW IT WORKS

Mains voltage is reduced by the transformer, and then rectified by the diodes D4-5 (4-7). The resulting d.c. is then smoothed to a degree by C1 and flows via R1 to light l.e.d. 1. This d.c. also ensures that pins 1 and 2 of IC1a (used as an inverter) remain high, so that its output pin 3 is low, preventing the thyristor from turning on. When the mains does fail C1 rapidly discharges with R1, I.e.d. 1, R2 and R3, and pin 3 goes high, providing gate voltage to the thyristor which then fires, allowing current to flow via the three bulbs. C1 has to be relatively low in value, otherwise it would take a long time to discharge, causing a delay before the light came on. S1 now acts as a simple on/off switch. When mains is restored the lights will stay on, because once fired, gate voltage can be removed from a thyristor. However, briefly opening S1 disconnects the anode from the battery which does reset the thyristor ready for the next failure.

When S2 is closed current flows through R5, which acts as a current limiting resistor to protect the batteries. D1 prevents battery voltage from reaching pins 1 and 2, as does D2. The purpose of D2 and D3 is to provide voltage to the supply pins of the i.c.s. Because the rectified d.c. voltage is higher than the battery voltage, D3 is normally reversebiased, so no current flows out of the battery. However, in the event of a mains failure D3 is now forward biased, maintaining supply to the i.c.s.

The closing of S2 also gates the slow oscillator constructed around IC1c and d. The frequency is adjustable to a certain extent by VR1. The output clocks IC2, a ripple counter, and after 16384 pulses pin 3 goes high, lighting l.e.d.s.

CONSTRUCTION

Assembly of the p.c.b. is straightforward. The plastic case used is the widely available Pifco "Everywhere Wall Light". It is first necessary to modify this case in the following way: The large bulb holder must be removed by drilling out its mounting rivets. Next, drill out the cord operated on/off switch. Drill two holes to accommodate the miniature toggle switches (see Fig. 1). These holes should be one above the other in the side of the case, directly opposite the "cord hole". This cord hole will be used to admit the mains cable. Two holes should also be drilled into the lens cover to take the l.e.d.s and (preferably) their bezels. Similarly, two further holes will need to be drilled through the internal diffuser to pass the l.e.d. wires (these wires should be fairly lengthy for lid removal). Hole diameters are dependent on switch and l.e.d. sizes, the former being subminiature types.

Only three nickel-cadmium cells are to be used, the mains transformers taking up one of the former battery positions. Various small transformers of the type needed are frequently seen advertised for as little as £1, therefore this component has not been rigidly specified. For this reason the rectification system has been made versatile to accommodate differing types, but the space available and the dimensions of the transformer must be taken into account. A little more spatial



★ Components

Resistors

R1, R2, R7	1k (3 off)
R3	10K
R4	4k7
R5	10 2 watts
R6	10M
R8	100k

All resistors 1/4 watt 5% unless otherwise specified

Potentiometers

4M7 VR1

Capacitors

C1	10µ 16V electro.
C2	470n polyester
C3	100n polyester

D1	1N4001
D2, D3	1N4148
D4-5 (4-7)	1N4001

D3	1N4148 (2 off)
5 (4-7)	1N4001 (2 off)
7	I.e.d.s (2 off)
	4011
	4020
	C106

Miscellaneous

D6-IC1 1C2 SC1

3 bulb holders (cat. No. PLH2 Home Radio) 3 torch bulbs 3V5 3 x 4AL V2 equiv. ni-cads 1 SPST) miniature toggles p.c.b. Transformer 7-0-7 Volts at 200mA Pifco Everywhere Wall Light, Model No. 1644

Constructors' note

The MES bulb holders are available from: Home Radio Components, 240 London Rd., Mitcham, Surrey, CR4 3HD. They may be second-sourced from: Electrovalue, 28 St. Judes Road; Englefield Green, Egham, Surrey, TW20 OHB. (Order skeleton lampholder type MSS36.)

The transformer's electrical requirements are less critical than its physical dimensions. For example, a 6-0-6V 250mA transformer will suffice so long as its measurements do not exceed 55mm wide, by 40mm high, by 35mm (the latter dimension to include any tags etc.).

If a 0-6V or 0-7V single-winding secondary transformer only, can be found, then the bridge rectifier arrangement may be used (insert D8 and D9 on p.c.b.). The Pifco lamp is available in high street shops.

> **Pifco Wall Light before** modification, showing the lens cover and its removable diffuser/ reflector.

Emergency Light (prototype)

C3

RR 100k

100 n



flexibility will be allowed if the two contact springs originally intended for the battery, are snipped out. Be careful how this is done. Do not remove the part which provides anchorage!

The three skeleton bulb-holders should be formed to the shape illustrated, and fixing holes drilled through the battery base-plate for mounting these. This shape has been conceived to aid the retention of the heavy nickel-cadmium cells. The round hole at the centre of the diffuser will not be large enough to clear the three MES bulbs in our new system, and therefore should be opened out using a knife or abra-file. It may be found necessary to bore some ventilation holes in the vicinity of the transformer to prevent gradual discoloration of the plastic.

When wired, the bulbs should light if S1 is closed. Switch on the mains, and the light should stay on; however, briefly opening S1 will extinguish them. Ensure that S1 is left in the closed (on) position, turn off the mains and the light should come back on.

To check that the batteries receive charge when S2 is closed, the most simple technique is to push a small piece of double sided p.c.b. in between the +ve of the battery and the battery holder. Attach pulses from a current meter (black to the +ve battery side because charge should be flowing from the more positive rectified d.c. side into the battery) and check that charging is taking place. Also check that when the mains is on S2 is open, that no current flows out of the battery (with the lights out of course!) To set the timer for 18 hours, monitor pin 7, which should change state every 63 seconds (63-27 secs-approx!)

CONCLUSION

One of these lights at the head of the stairs is a real safety feature at home. It makes an ideal present, perhaps particularly for elderly people who may find a sudden blackout more difficult to cope with.

The only component possibly to cause problems when purchasing the parts is the case. It is manufactured by PIFCO and is called the "Everywhere Light" (model 1644) though other cases may, of course, be used. Because the batteries are not being trickle-charged, it is good practice to check the state of charge every so often.

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The 'in' thing for absent residents

 $T(1) \cap (C) \cap T$

HERE is a light that will deter a burglar in your absence as it automatically switches on at dusk and off at dawn. As such, it is ideally suited as a porch light, giving the impression that the premises are occupied.

Whilst the supply requirements are low, the period when the unit is in operation is long, so the use of the mains voltage was considered imperative and anyway, if the light was not powered from the mains it would probably give the game away.

CIRCUIT

In Fig. 1, a mains transformer T1 is used to step down the 240V supply to the operating potential. The secondary winding supplies the rectifier pair D1/D2 to provide rough d.c. which is further smoothed by the electrolytic capacitor C1.

TR1 is used as a switching transistor having the relay coil as its load. The diode strapping this is to suppress the transient high reverse voltage which would normally develop when the relay is switched off. Its action is to simply short circuit the high voltage spike, which could damage the transistor, reducing it to a value of about half a volt.

Bias is provided to the switching transistor by a resistive chain which includes a cadmium sulphide light dependent resistor, the resistance of which reduces as the light falling



EG 802

Fig. 1. Circuit of Autolight



Fig. 2. Veroboard layout

on the device increases. Typically, for an ORP12 this resistance ranges from 10M in total darkness to less than 100 ohms in bright conditions, and around a few kilohms for a cloudy day.





A sensor that sounds for liquid or vapour

LIQUID ALARN

ONE of the simplest and most versatile of direct coupled circuits is the super-alpha or Darlington pair. Super alpha derives from the gain multiplication produced by connecting transistors this way, so that the composite pair shown in Fig. 1 have a very high current amplification which for the transistors being used approximates to 125^2 .



EG794

Fig. 1. Circuit of Alarm

VEROSWITCH

A piece of 0.1in matrix Veroboard with the alternate track layers commoned (Fig. 1) makes an ideal sensor as liquid or vapour bridging these will form a base input to TR1.

In the initial setting up, the sensitivity of the device is set by VR1. If steam is applied to the parallel copper strips of the sensor, VR1 is rotated until the alarm sounds. At this point the adjustment is for maximum sensitivity.

In its appointed role as a laundry protector, obviously the housewife will only be concerned with rain that has started falling so that she can make the quick rescue, so a high sensitivity adjustment should be made.



Fig. 2. A mains supply that can be used

LIQUID LEVEL

If a liquid level switch is intended, where two stiff conductive probes are substituted for the Veroboard, then the simple expedient is to use lengths of 16 s.w.g. tinned copper wire for the probes, these being terminated with the input connections to the device.

The retainer for these stiff wires could be a five-way terminal so that adjustment for probe separation, and therefore liquid level, would reduce to loosening the screws of the block and moving the wires relatively as required.

To complete the assembly, the two outer connectors of the block could retain a suspension hook of wire.

Obviously the probes could be manually adjusted for any liquid level rise or simply left with the tips adjacent to detect a level.

ALARM

The choice of supply was deliberately kept low for battery conservation, as the current consumption of the alarm rises almost linearly with the voltage applied. For the circuit shown and the alarm arranged for pulsed tone operation (3.2kHz approximately pulsed at 5Hz), the current is an intermittent 10ma. Nevertheless, sound output is strident and will almost certainly be heard anywhere around the house. However, if you do need a bigger output then you can simply raise the supply voltage to 9V. There are no circuit changes to be made, but remember, whilst the db output level rises so does the current consumption.



Fig. 3. Veroboard layout

OUTPUT ALTERNATIVES

With the alarm specified the output alternatives are continuous or pulsed tone depending on whether the yellow sense lead is connected. The instruction for this is to leave it unconnected for pulsed operation or to connect to battery negative for a continuous tone. The remaining two leads (red and black) are connected as shown in the circuit.

With its obvious long-term utility there is a need to substitute a simple mains supply for batteries and a suitable 9V circuit is given in Fig. 2.

COMPONENTS

Resistors

R1 10% ¹/₄ watt carbon

Potentiometer VR1

Transistors

TR1-TR2

10k preset

BC171A (2 off)

1k



Warning Device WD1—Surface mounting audible alarm (RS 248–788)

Battery B1—see text

DOUBLE SIREN



Alarm options with one transistor

THIS project makes use of only one of the free transistors but offers two siren options: an American police siren with the characteristic short rise and longer fall and alternatively the urgent rising bursts of sound reminiscent of the alarm status signals of the starship Enterprise. It should be noted that although the component count is small the sound output is large.



EG 7 9 1

Fig. 1. Circuit of Siren

BLOCKING OSCILLATOR

Both these sounds are initiated by S1 and S2 respectively, but for the starship alarm S1 must be kept closed while pressing S2. Power consumption is relatively low---about 24mA---but if batteries are heavy on the pocket then there is the simple expedient of knocking up the simple 9V power supply given previously in this supplement.

This circuit has the rather formidable description of a blocking oscillator. Here, the transformer provides regenerative feedback by phase inversion from the collector to the base of TR1; so remember, if your unit does not work just reverse the connections to the primary or secondary of T1.

When S1 is pressed the tone produced through components C1, T1, R1 and R2 is raised by the charging of C2 through R3. When S1 is released the charge in the electrolytic capacitor leaks away so that the tone gradually falls. By introducing the diode with the switch S2, C2 is very rapidly discharged so that with operation of this switch you only have a series of rising tones.

OUTPUT

In introducing this project it was stated that the sound output is large. This statement is true if a large speaker is used, anyway, you can have fun experimenting. I used a 4in type and found the output adequate.

Since sound expansion has been touched upon there
COMPONENTS Resistors R1 R2 R3 All 4 W 10% carbon	10k 18k 3.3k		
Capacitors C1 C2	100n 500µ 25∨ elect	Transistor TR1	BC171A
Switches S1–S2	Press switches	Loudspeaker LS1	4in 8Ω
Transformer T1	T/T7 output (RS)	Battery B1–9V	



could possibly be a requirement for changing the decay period or frequency. Although I have endeavoured to simulate the sounds set out in the introduction, I certainly am not gifted with perfect pitch so there might be some among you who require a little more accuracy in the reproduction, then the guidelines for sound tailoring are that increasing R1 will increase the siren's fall period and increasing C2 will lower the frequency and, of course, vice versa.

AF/RF GENERATOR



Broad band signal source

THIS little generator makes no pretentions towards a high specification, but with it you could quite easily troubleshoot a radio covering the medium and long wavebands, which includes the r.f. and a.f. components, or any audio entertainment equipment.

FREE RUNNING

The circuit shown in Fig. 1 is a free running square wave generator which produces a basic tone of frequency 1/(1.4CR) cycles where C is the single coupling capacitance (farads) and R the base bias resistance (ohms), which works out to 1 kHz.

To reduce signal output the simple expedient of starving the circuit is used via VR1.

ACTION

When power is first applied to the circuit from the battery, since there is no means of maintaining a permanent voltage at the base of either transistor the circuit has no stable state. The unbalance between each half of the circuit induces one transistor towards current cut-off and the other towards full conduction. However, there is no d.c. stable state and the current alternates between the two quasi-stable states at the previously mentioned rate fixed by the 200k/5n pairs. As the value of these are equal the output forms a symmetrical square wave.



Fig. 1. Circuit of Generator

HARMONICS

The harmonics produced to cover the r.f. bands mentioned are the normal components of a square wave. How far they extend upward being set by the characteristic of the wavefront. The shorter the time of the rise the higher the harmonics extend upwards.

OUTPUT

Since the transistors are switched hard on the output voltage is that of the supply. To vary this output it is convenient to vary the supply with the potentiometer VR1, since the current drawn from the battery is extremely low. In fact, the generator will work with nothing more than a meagre 0.5V.

The output coupling capacitor C3 has a working voltage which has been chosen deliberately high for any valve circuit applications. Make sure the signal lead from this is coax suitably terminated with crocodile clips.

USING IT

Assuming you have no test equipment, when applying the



Fig. 2. Showing board assembly

generator to entertainment equipment fitted with its own loudspeaker the rule is always to work back from the output stage to the input. The evidence of stage and then component failure in proceeding in this logical manner is loss of signal.

With, say, a radio, this should be tuned off a station with the volume turned up. You should then work back through the a.f., i.f., and r.f. stages to locate the faulty component. A good set will produce a loud tone if the generator output is brought close to the ferrite aerial.

COMPONENTS Resistors R1 R2 R3 R4 ·All $\frac{1}{4}$ W 10% carbon	10k 200k 200k 10k		
Capacitors C1-C2 C3	5n 5n	Transistors TR1–TR2	BC171A (2 off)



Touch tournament with a memory

ERE is a constructional variation of the old 'steady hand' game where a loop of wire has to be run along another piece of wire bent into an irregular pattern. The object of the game is to run the loop from one end of the shaped wire to the other without making contact between them. If this is done, then the loop is returned to the start and the next competitor makes his move.

Since any number of people can compete and presumably complete the course, the winner must be the one who does it in the shortest time.

MORE DIFFICULT

The difficulty of the game can be improved upon by any



Fig. 1. Circuit of Game

one of the following changes, which are a) making the curves of the wire length more irregular, b) lengthening the run, and c) decreasing the diameter of the loop. These are all variations which can be tailored to suit the age and skill of the competitors and which cost very little in time or money to arrange.

HOW IT WORKS

The circuit shown in Fig. 1 is that of a compound emitter follower which has a characteristically high input impedance seen from the base of TR1 and the cathode of D1.

If, say, there was a short between the competitor's loop and the shaped wire the light emitting diode would switch on and in the absence of the electrolytic capacitor C1 this would switch off as soon as contact was broken.

You can imagine, then, the argument that would ensue if the offending party refused to own up to the instantaneous flicker, so C1 was introduced to remove any conflict.

In practice, the l.e.d. will remain lit for periods in excess of five minutes for the slightest touch so it does arbitrate conclusively. The switch S1 discharges C1 in preparation for a new game.

CONSUMPTION

The unit consumes no current in the stand-by condition and 40mA when switched on. This is, in fact, the maximum forward current of the l.e.d. specified and if a battery is to be used some might like to conserve power with a reduction of light emission.

To do this, resistors from the one specified up to a value of 1k can be used. The 1k limit provides a visible lamp but considerably reduces the consumption to about 5mA.

POWER SUPPLY

Once again we can solve the problem of battery cost by building the p.s.u. shown in the Liquid Alarm project.



Fig. 2. Board assembly

COMPONENTS		
	6	
• • • • • • • • • • • • • • • • • • •		
Resistor R1 All 1 W 10% carbon	120	
Capacitor C1	100µ 15V	
Transistors TR1–TR2	BC171A	
Diode D1	0-2in g re en I.e.d. (RS 586–481)	
Battery	B1 —9V	

CAR BATTERY CONDITION INDICATOR



MOST car manufacturers pay scant attention to dashboard indicators in the charging department. A red light that comes on and goes off when you start the journey leaves you electrically ignorant in its absence until something like a broken fan belt occurs.

Lights the way to happy motoring

Just over a generation ago the voltmeter and ammeter were standard equipment for the motorist so that he was constantly aware of what was happening to the battery, dynamo and regulator when the vehicle was started or running.

Today's economies reduces all electrical awareness of a satisfactory circuit to the standard ignition warning light, which, apart from a perfunctory introductory glow, tells you nothing about the battery state of charge.

FULL INDICATION

This three l.e.d. alternative will continuously monitor the general state of the battery from 'poor' (red l.e.d. lights), when suspicion should be cast on the battery voltage regulator or dynamo or combination of these. For this condition the battery voltage is below 11.6V.

A battery terminal voltage above this would light D4, the yellow l.e.d. This condition warrants an investigation if it does not charge up after a few minutes and is marked 'fair' on the circuit.

Indication of a satisfactory circuit with a battery voltage of above 12.6V is given by the green l.e.d. lighting. This is labelled 'good' in Fig. 1.

There are transitional states between these as when the yellow lamp reduces in brilliance and transfers to a brightening green lamp, as a battery is charging, and vice-versa as it is discharging.



EG791)

Fig. 1. Circuit of Indicator

IN GENERAL

Broadly, the green light on indicates all is well; red, the battery is defunct and yellow the output voltage is low which will be confirmed by poor light from the lamps, low hydrometer reading and lack of power to the starter motor when the battery is switched in.

This condition may be due to the dynamo not charging or giving a low or intermittent output in which case D4 will flicker.

POSSIBLE REMEDY

To remedy this latter condition the charging and field circuit wiring should be examined. Tighten any loose connections, particularly those at the battery, also renew any damaged cables. Check the tension of the dynamo driving belt also the regulator setting and adjust if necessary.

MOUNTING

The small board can be mounted directly on the car dashboard using the black panel clips supplied with each l.e.d. These suit a 6.35mm (0.25in) panel and will retain the circuit board when fitted.

A layout with equally spaced light emitting diodes is shown in Fig. 2. The viewing angle for the specified l.e.d.s is 30° which might produce a relatively low light output in bright conditions, so I am proposing an alternative device, the 'stackable l.e.d.' from RS Components, which has a viewing angle of 100° and provides a clear colour contrast when viewed at even oblique angles.



EG796 A 16796 8

Fig. 2. Veroboard layout

COMPONENTS			
Resistors R1 R2 R3 R4 R5 All ½W 10% carbon Semiconductors D1 D4 D7 D2. D5 D3 D6 TR1, TR2	390 1k 390 1k 390 1k 390 Red 0.2in l.e.d. (RS 586–475) Yellow 0.2in. l.e.d. (RS 586–497) Green 0.2in l.e.d. (RS 586–491) IN4148 (2 off) 11V 500mW Zener 12V 500mW Zener BC171A (2 off)		

ANOTHER ADVANTAGE

An additional advantage of this indicator is that being 'in circuit' continuously means that the battery condition can be tested as different loads are applied such as operation of car lights, starter motor, accessory items, etc. This way one can readily keep an eye on a deteriorating battery and know the time for replacement rather than be caught unawares.

CIRCUIT ACTION

In the circuit, at voltages below 11.6V, the brightness of the l.e.d. D1 is set by R1. As 11.6V is approached then D4 starts to light as TR1 conducts, the latter subtracting current from D1 causing it to extinguish.

This same procedure is repeated as the line voltage moves up to 12.6V when D7 will be lit. So you can see that there is always a continuous visual indication of battery state, and in living with it you should become extremely adept at interpreting impending problems.

CHECKING IT

When checking over the completed unit a variable power supply is useful to make sure the unit is switching through its ranges.

If dry batteries are used for this purpose it should be borne in mind that for a 'good' indication the total current drawn is around 80mA, which is largely made up of shunt current through D5.

The l.e.d.s chosen for this project are for panel mounting and could be attached to the car dashboard but first you should view these in daylight when the unit is powered up (take the 12V line from the ignition switch) for optimum positioning as brightest viewing is directly over the lens.

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ULTRASONIC VISION SYSTEM

This project is an advanced ultrasonic vision system which is within the means of the amateur constructor, yet is sophisticated enough to enable experimentation with computer measurement, image recognition, object tracking, and robot vision.



The PE Quasar stereo cassette deck is a versatile low cost system designed around a single p.c.b. which features a variable bias control for optimum recording of all tapes, a gate noise reduction system, twin VU meters and electronic switching.

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IR REMOTE CONTROL Part 2 Bart Trepak

HE transmitter circuit is built on a small printed circuit board as shown in Figs. 1 and 2 which holds all the components except the l.e.d.s and the push switches. Care should be taken to ensure that transistors and electrolytic capacitors are mounted the correct way around. Although the i.c. is a bipolar device and, therefore, not prone to static damage, it is wise to mount this in a socket. If only one switch is to be used, a panel mounted push to make type may be fitted directly on to the box and connected to the appropriate pads on the transmitter board by wires. If more than one switch is required, these should be mounted on the switch board (Fig. 3), which is designed to fit "piggy back" style on the transmitter board using four stiff wires (pins). Care should be taken to ensure that the switches are positioned as shown in Fig. 4 and firmly seated on the p.c.b. before soldering.

The complete circuit may be housed in a small box together with the battery. The l.e.d.s should be mounted on the front of the box using l.e.d. clips or the reflectors if a greater range is required. If the box specified is used then these would be mounted on the aluminium front panel and soldered to the p.c.b. using flexible wires. The wires to the l.e.d.s should be kept as short as possible as relatively large



currents will flow in these leads and any inductance will have a detrimental effect on the range. Since there are no heavy components on the p.c.b., the transmitter circuit may be secured to the box with a double-sided adhesive pad. The distance between the main p.c.b. and the keyboard will have to be carefully chosen to ensure that the switches are free to move when the box is clipped together. If the assembly is made too high, one or more of the switches may be depressed permanently, which would, of course, result in the transmitter operating continuously. When the correct height has been ascertained, the connections between the two p.c.b.s may be made using tinned copper wire or discarded component leads. The two p.c.b.s have been designed so that the holes for the interconnections line up when the two assemblies are mated. Only the connections for which holes are provided should be made, the spare holes on the main p.c.b. (which are for use with other keyboards) should be ignored.

The battery clips should be assembled in the box battery compartment and connected to the p.c.b. pads marked + and — with wires. If there is any danger of incorrect battery connection, it is wise to place a diode (type 1N4001) in the positive lead to avoid damage. The circuit is powered by **a** 9V PP3 battery and draws an average current of 20mA when operating. Since the standby current consumption is only a few micro-amps, no on/off switch is required. A piece of polystyrene or similar material may be required to hold the battery in place within the battery compartment.



Switch assembly





Fig. 1. P.c.b. design for the transmitter

Fig. 2. Component layout





Fig. 4. Switch layout

CONNECTIONS TO MK8 KIT

...

0 2

00100

ONI

00010

SERE SERE

EG765



Fig. 5. P.c.b. design for the Receiver board



Fig. 6. Component layout

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The receiver is built on one printed circuit board as shown in Figs. 5 and 6 with all the components, including the transformer, mounted on it. (Construction on Veroboard or other p.c. layouts is not recommended as the high gain amplifier IC1 may become unstable.) Only D1 and the opto isolator and triac are mounted remotely. D1 may be connected to the p.c.b. with wires although the length of these should be kept as short as possible to avoid noise pickup. A screend lead is best although the length should be kept to less than 6 inches as any capacitance across the photodiode will reduce the sensitivity of the circuit to the transmitted IR pulses. As the photodiode will need to be mounted externally, some form of protection from the weather will be needed.



The sensitive face of the diode is the curved one and this should be arranged to face in the direction from which the IR radiation is expected. Any transparent or translucent enclosure of the type supplied for panel mounted neons would be suitable provided the diode, is mounted against the front of the moulding to avoid making the unit too directional. If the unit is to be used in a room to control lamps or appliances, the photodiode mounting will not be so critical as multiple reflections occur from walls and ceilings making the controller largely non-directional.

The unit may be tested by connecting the mains to terminals J and K and transmitting the open/close code. VR1 should be adjusted until l.e.d. D2 lights and remains lit while the transmitter button is pressed. The other codes may be tested by connecting l.e.d.s directly to terminals E F and G H, with the l.e.d. cathodes connected to E and G. Transmitting ON 1 and ON 2 will switch each l.e.d. on while pressing OFF will cause both l.e.d.s to switch off.



Fig. 7. P.c.b. mains switch

Fig. 8. Component layout

The layout of the mains switch is shown in Figs. 7 and 8. The triac CSR1 specified is an 8 Amp device which can control up to 2kW at 240V a.c., although a suitable heatsink will be required to keep the triac case temperature to below 70°C. Without a heatsink, a load of up to 1.5A (300W) may be switched. If higher power loads are to be controlled a TIC236D (12A r.m.s.), TIC246D (16A r.m.s.) or TIC263D (25A r.m.s.) may be fitted without any circuit modifications. Note that the load must be connected in series with terminal Lo. Since the input is an I.e.d., the correct polarity must be observed with terminal M connected to the positive line, i.e. F or H.

APPLICATIONS

The obvious application for the unit and indeed the purpose of the design is in the control of motorised garage doors. In this context the momentary output would be used to trigger the open or close sequence while the two latched outputs could control mains lamps (via the slave switches) to illuminate the drive and garage if required.

The circuit could of course be used for a multitude of other functions around the house. As mentioned, the ON/OFF outputs could be used to switch virtually any appliance on or off such as radios, tape recorders, lamps, etc. The momentary output controlled by the open/close button could be used to change the channel on a suitably modified TV receiver. An interesting application would be to control room lighting. Many modern touch controlled lamp dimmers incorporate an input for connecting to slave touch plates. These normally require a momentary low current connection to the live terminal to effect switching or dimming of the light.

For the elderly or disabled, the unit could be used to control heating or lighting. Since the connections to the power switch of Fig. 4, shown last month, need only to be low voltage/low current wires, it would be easy to arrange for the controller to be situated in one room and control an appliance in another location, enabling, for example, the central heating or an electric blanket to be switched on or off without the necessity of leaving the living room.





Michael Tooley B.A. David Whitfield M.A.M.Sc. PART THREE

THE purpose of the S-meter module is to provide a d.c. output proportional to the amplitude of the 455kHz IF signal. The module samples the level of the IF signal in the Ranger preceding the amplitude limiting stages within IC100. At this point (the output of FL100) the IF signal is only at a relatively low level and hence some amplification is essential before detection. To prevent instability the module must be housed within the Ranger transceiver and its output taken via the 7-way ancillary connector to the Base and Mobile Adaptor. Where signal strength indication is not required, the S-meter module may, of course, be omitted. In this case there will simply be no meter indication on 'receive' but otherwise the performance and operation of the unit will remain unchanged.



Fig. 3.1. Circuit diagram of the S-meter module

The circuit diagram of the S-meter module is shown in Fig. 3.1. Source follower, TR1, provides impedance matching and its high value of input impedance minimises loading of the IF amplifier at the point where the signal is derived. This is an important consideration as the filter needs to be accurately matched in order to obtain optimum performance. TR2 and TR3 form a voltage amplifier with gain adjustable by means of VR1, D2 and D3 rectify the amplified IF signal and the supply voltage is held constant by means of D1.

The S-meter module is assembled on a small single sided p.c.b. for which the foil layout is given in Fig. 3.2. The corresponding component layout is shown in Fig. 3.3. The

completed p.c.b. should be wrapped in a polythene sleeve and located at right angles to, and on the right hand side of, the main p.c.b. in the area adjacent to IC100 and IC101. Only four connections need to be made to the Ranger and these should again be kept as short and direct as possible.

DDDDDDD



Fig. 3.2. P.c.b. design for the S-meter module



Fig. 3.3. Component layout for the S-meter module

The input is taken from the junction of R111 and FL100. This connection may most conveniently be made by soldering to the left-hand end of R111 (the end nearest FL100) on the top side of the p.c.b. The positive supply is derived from test point G and the common rail is taken from pin-2 of SK203 (or any other convenient OV point). The output of the module is taken to pin-1 of SK203 and note that, regardless of whether the S-meter module is fitted or not, some minor modifications to the wiring of SK203 are necessary and these are covered in the next section.

TESTING AND ALIGNMENT

During the test and alignment procedure it is recommended that a current limited regulated d.c. power supply be used. The supply should be set to give $12V \text{ d.c.} (\pm 0.5V)$ and its current trip set to $1A (\pm 100\text{ mA})$. The Base and Mobile Adaptor does have its own internal 1A fuse but this should not be relied upon in the initial stages. Protection is essential in order to minimise the risk of damage to both the external power supply and to the Base and Mobile Adaptor in the event of incorrect component connection or faulty assembly (e.g. solder bridges between p.c.b. tracks). Under no circumstances should the external supply be allowed to exceed 17V as this may result in permanent damage to both the Ranger and the Adaptor.

Two interconnecting leads are required. Both need be no more than 160mm long since it is envisaged that the two units will normally be operated 'piggy-back' rather than 'side-by-side'. Constructors should, however, make the connecting cables long enough to allow the two units to be operated 'side-by-side' during the setting-up process. The RF connecting lead (at the front of the set) should use good quality 50ohm coaxial cable. (e.g. UR43 or UR76) terminated with PL259 plugs. The correct method for fitting these plugs is shown in Fig. 3.4. and 3.5. The control lead (at the rear of the set) has seven ways and should be terminated with 7-pin DIN plugs. Care must be taken to ensure that the pins are correctly linked and this is best done by checking pins 1 to 6 with an ohm-meter both for continuity and for the presence of short circuits between adjacent pins.

Make the following adjustments and connections in the order given before connecting the external d.c. supply:

- (a) Ensure that the interconnecting leads are in place.
- (b) The channel selector should be switched to channel 14.
- (c) The 10dB attenuator should be switched 'off'.
- (d) The RF gain control should be set to maximum (fully clockwise).
- (e) VR1 should be set to minimum resistance.
- (f) The cores of L1/L2 and L3 should all be set two full turns from the furthermost position into the former.

COMPONE	NTS
S-Meter Module	
Resistors R1, R4 R2, R5 R3 R6 R7 All resistors $\frac{1}{4}$ W 5%	100k (2 Off) 2k2 (2 off) 4k7 1k5 1k 6 Carbon.
Capacitors C1 C2 C3, C4	22p 1n 4n7 (2 off)
All capacitors cerar Semiconductors D1 D2, D3 TR1	BZY 88 C5V6 OA91 (2 off) 2N 3819
TR2, TR3 Miscellaneous VR1	BC 548 (2 off) 220 min preset skeleton (hor)

- (g) The variable capacitors should be set as follows: VC1 and VC2; plates 20% meshed. VC3 and VC4; plates 50% meshed.
- (h) If the filter modules have been fitted they should be temporarily removed and replaced by shorting links.
- (i) Connect an external loudspeaker (of between 4 and 16 ohms impedance) to the Adaptor via SK3.
- (j) Switch the 'squelch' on the Ranger 'off'.
- (k) Connect a matched 50ohm aerial to the Adaptor.

The external d.c. supply should now be connected and the steps which follow should permit alignment of the Base and Mobile Adaptor:



STEP NUMBER	PROCEDURE	NOTES
1	Switch 'on' the external d.c. supply. Switch 'on' the Ranger and the Adaptor unit.	The 'receive' I.e.d. indicator on the Adaptor should be illuminated and noise should be heard from the Adaptor's loudspeaker when the volume control on the Ranger is advanced
2	Adjust the core of L1/L2 for maximum indication on, the S-meter using the signal from a fairly strong local station. If a full-scale reading has been obtained reduce the setting of the RF gain control and continue to adjust L1/L2.	If the S-meter is not fitted adjust for minimum background noise (maximum quieting).
3	Select a channel on which there is a weak but stable signal. Adjust the RF Gain control to provide a reading of about $\frac{1}{3}$ to $\frac{1}{2}$ scale. Re-peak L1/L2 for maximum.	Only a slight adjustment, if any at all, should be necessary.
4	Disconnect the RF (coaxial) cable linking the Ranger and Adaptor. Connect the aerial directly to the Ranger thus bypassing the Adaptor. Note the S-meter reading obtained when a weak-but stable signal is present.	
5	Re-connect the Adaptor and note the new S-meter reading with the RF gain control turned to maximum. (The same signal should be present as in step 4.)	If the RF stage has been correctly aligned and is working properly there should be an increase of approximately two S-units (around 12dB). If no discernible change is obtained or if there is a reduction in signal strength check the RF stage and repeat steps 1 to 5.
6	Replace the aerial with a 50ohm resistive dummy load capable of dissipating 5W. (Such devices are available from most CB accessory stockists.) Operate the Ranger on 'transmit' by either using the 'press-to-talk' facility on the microphone or by using a shorting link as described in the Ranger alignment procedure.	The 'transmit' indicator I.e.d.'s should be illuminated on both the Ranger and the Adaptor, and the noise from the loudspeaker should cease.
7	Adjust VC1, VC2, L3, VC1 and VC2 in the sequence given for maximum output power indication from the meter. Repeat several times and adjust the setting of VR1 to reduce the indication on the meter if necessary.	The adjustments are fairly critical and there is interaction between them.
8	Adjust VC3 and VC4 for maximum meter indication. Repeat several times for maximum output. Again adjust VR1 if necessary.	These adjustments are faily broad with some interaction between them.
9	If an RF power meter (or SWR bridge which incorporates power measuring facilities) is available, check the output power delivered to the load. This should be between 3W and 6W. (See Fig. 3.6.)	If this is not the case repeat steps 7 to 9 inclusive.
10	Insert the filter modules and carefully adjust the ferrite cores for maximum output. Some minor re-adjustment of VC3 and VC4 may be beneficial.	The output power will be somewhat less than that obtained previously and should be typically between 2W and 5W.
11	Adjust VR2 in the Ranger for an output power of 4W.	Slight adjustment to the output stage and/or filter in the Ranger may help improve the drive level if necessary.
12	Switch the 10dB attenuator 'on' and note the effect on the output power. This should fall to approximately 400mW.	If the output power is greater than 500mW the value of R9 should be increased to 100 ohm 2W. If the output power is less than 300mW the value of R9 should be decreased to 47ohm 2W.

This completes the setting-up procedure and the Adaptor is now ready for 'air-testing'.

Testing and Alignment procedure



Final plug assembly with coupling ring free to mate with a socket.

Fig. 3.5. Method of fitting PL259 connector to RG 8/U, UR67 or similar 0.4in diameter co ax cables.

AERIAL MATCHING UNIT

There will doubtless be many occasions where, for one reason or another, it is not possible to achieve a good match between the transceiver and an aerial system. Usually this is because the radiating element is not cut to the correct length; either it is too long or too short. Aerials with adjustable radiating elements are obviously to be preferred but they can be tiresome and inconvenient to adjust, particularly when they are mounted at some considerable distance from the rig. A most useful accessory, therefore, is a device for accurately matching the transceiver to any aerial system. The effect of providing an accurate match between the transceiver, feeder and aerial is to guarantee a low value of SWR. This helps to improve the radiation efficiency by eliminating additional SWR losses in the feeder and maintains the performance specification of the transceiver itself. The matching unit consists simply of a pi-network of two variable capacitors and a fixed inductor. An added bonus is the reduction of harmonic radiation associated with the lowpass characteristic of the network.

The circuit diagram of the aerial matching unit is shown in Fig. 3.7 together with an internal layout and wiring diagram



£1 Fig. 3.8. Internal layout of the aerial matching unit



EA324



сомро	NENTS
Aerial Match	ing Unit
SK1, SK2	S0239 sockets (2 off)
VC1, VC2	Air spaced variable, max. capacity in the range 365 to 500pF, see text
L1	8 turns 18 s.w.g. enamelled copper wire, inside diameter 10mm, winding length 10mm approx.
Metal case or	diecast box
Pointer knobs	(2 off) fitted with '0-10' scale, see text

in Fig. 3.8. A metal screening enclosure is essential to the correct operation of the matching unit and this can take the form of a diecast or aluminium box measuring approximately 150x80x50mm. The variable capacitors may be difficult to obtain but fortunately their exact maximum value is unimportant and can be anywhere between 365pF and 500pF. The prototype employed dual gang 176 and 208pF capacitors with the two gangs connected in parallel to make a total maximum value of approximately 384pF. Capacitors of this type were commonplace in AM (medium and long wave) transistor portable radios of some 10 to 20 years vintage. Whatever type of capacitor is used they MUST be air spaced types and should be fitted with a pointer knob and scale. This is important for providing accurate reference points for re-setting the matching unit when it is to be used with several aerial systems. The unit should preferably be fitted as close to the aerial as possible. This is particularly important when a high SWR (greater than 2:1) is present. In many situations, however, this may not be practicable and it will be necessary to place the unit in the vicinity of the transceiver and SWR bridge. This is a less efficient arrangement (due to the high SWR in the feeder) but it will at least provide a reasonable match for the transceiver. A typical arrangement is shown in Fig. 3.9 and the aerial matching unit is simply adjusted for minimum VSWR. In practice, and with careful adjustment (both controls need to be used) it is possible to obtain an SWR of 1.1:1 or better with even the most difficult aerial systems.

Digital Design Techniques...

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Part 8 Memories & Microelectronics

N the series so far we have looked at the design of logic systems, from simple combinational and sequential circuits up to multiplexing and bus-orientated networks. It's only a small step now into the world of microcomputing and microprocessors. We need only one more 'building block' of circuitry; the memory.

BASIC REQUIREMENTS OF MEMORY

Essentially, 'memory' is a collection of logic 1's and 0's which are stored in a system. There must be some way to 'write' these logic states into the memory, and some way to 'read' the logic states out of the memory and into other circuitry.

In some cases it might not be necessary to write information into the memory more than once. If a fixed set of binary numbers was to be stored, always the same and never requiring modification, then a 'Read Only Memory' or 'ROM' could be used; in its simplest form this would be a set of wire links or switches permanently set to give a logic 0 or logic 1 where required. In many cases, though, it is necessary to be able to frequently write information into the memory, as well as being able to read it. Such a memory is known as a 'RAM', which stands for 'Random Access Memory', meaning that all bits stored in the memory can be directly and independently written into, or read from. 'RAM' is actually a bad name for the device; all bits can be accessed randomly in a ROM, too. Unfortunately, this is a terminology that we'll have to live with!

In both ROMs and RAMs we want to be able to choose a specific point in the memory, known as the "address', and read (or write) the information present there into the rest of our system. The memory device, therefore, will have a number of binary inputs which together form the address, and a number of binary outputs (or inputs/outputs in the case of RAM) which together form the data output (or input). Each value of the binary number on the address inputs is known as an address 'word'. In the case of RAMs, there is a read/write control input, and in many devices an 'enable' input is provided, to allow all functions of the device to be turned on or off. The timing of the logic signals can be very important; in many memory systems there must be certain minimum time periods allowed between data changing and the logic state of the read/write control being changed. Failure to observe this can cause incorrect data storage and retrieval, and is an important consideration when designing with many different types of memory device.

STORAGE DURATIONS

Systems which lose their information when the power supply is turned off are known as 'volatile', while those that

retain their information are 'non-volatile'. In some cases it is possible to use battery back-up supplies with a volatile memory to provide short term storage with the power off. In other instances it is possible to specifically design nonvolatile memories, although this is usually done at the expense of performance or memory cost.

Frequently, the use of memories can be broken down into 'bulk storage', and temporary or general purpose storage, often known as 'scratchpad memory'. The former is usually a non-volatile technology, often with relatively slow access or operating speed but capable of holding a massive amount of information for permanent records, storage, filing, etc. The latter is used on a 'day-to-day' basis by the circuitry it serves, mostly for short term storage of information, rather in the manner of a jotter or scribble pad on a writing desk. (Hence the term 'scratchpad' memory). In order to avoid slowing down the surrounding circuitry, this memory has to be a fairly high speed type, although it does not usually need to store a very large amount of information.

MEMORY TECHNOLOGIES

Memories can be implemented by using various different technologies; electromechanical devices (relays, uniselectors, and the like) being one of the simplest, although these are far too slow, bulky, and cumbersome for the majority of applications. In early computers the non-volatile 'magnetic core' was used. This consisted of a large number of small ferrite rings with wires passing through their centres. By passing impulses of current through selected wires, the ferrite rings could be magnetically saturated and hence these cores were written into. By passing current pulses of the opposite polarity into the selected wires, and measuring the resulting e.m.f., the saturation of the cores could be determined and the memory could be read. Although a widely used technique several years ago, the magnetic core has now all but disappeared. More recent magnetic technologies are those of tape, disc, and bubble.

Magnetic tape is familiar to us in the form of audio cassettes, and indeed these are often used in small 'personal' computers as bulk storage, with differing frequencies of tone being recorded in bursts to indicate logic O's and logic 1's as appropriate. Larger computers use larger tape; many times the thickness, width and length of cassette tape, running at very high speed, and with many tracks used in parallel to increase the speed of reading and writing of the memory.

Magnetic tape can store fantastic amounts of information, but is inherently very slow; if the data that you want is at the opposite end of the tape it can take several minutes to get there! A natural progression, therefore, was to use a magnetic disc. With this system, a circular disc coated with a thin ferromagnetic material (similar to that used on tape) is continuously rotated at high speed. In the large, solid, 'hard' disc units, the read and write heads float on an air cushion fractionally above the disc, whereas in the case of the smaller flexible 'floppy' discs, the head and disc actually touch, the disc being coated with a lubricating material to reduce friction.

Information recorded anywhere on the disc can be accessed in a fairly short time by moving the heads inwards or outwards along a radius from the centre of the disc, and then by waiting for the disc to rotate sufficiently to bring the required information directly underneath the heads. In this respect, the disc is analogous to an LP record; you don't have to play the whole LP to reach a track that you want to hear—the pickup can be placed on that track directly, manually, saving a great deal of time. This method of bulk storage is fairly fast, and can store large amounts of data; typically between 0.1 and 1 Megabytes (a 'byte' is a group of eight bits) for a floppy disc and over 10 Megabytes for a hard disc.

A fairly recent development in non-volatile bulk storage is the 'Bubble Memory'. This device consists of a very thin layer of garnet crystal, grown on a non-magnetic substrate, wrapped with flat wire coils, and enclosed within a pair of permanent magnets. A large number of geometric 'shapes' are deposited on the surface of the garnet, in a soft magnetic material. Tiny magnetic 'bubbles' can be generated in the garnet; these are small areas only a few micro-metres in diameter, in which the magnetic field is opposite to that in the rest of the crystal. Magnetic fields generated by the coils produce and control the magnetic polarisation of the deposited geometric 'shapes', which in turn manipulate the bubbles. The presence of a bubble represents a logic 1, and the absence of one a logic 0.

Bubbles can be sensed by monitoring the resistance of extra 'detector' shapes. As the bubbles pass under these detectors, the resistance of the deposited shapes changes, and this can be amplified by a suitable detection circuit. The circuitry needed to control a bubble memory is complex and expensive, and the whole technology at the moment is a rather specialised one. In its favour, it is fairly compact, nonvolatile, low power, and has reasonable speed performance. Of all the memory technologies in current use, it has perhaps the greatest potential for improvement and developments.

The memories that we've looked at so far have had attractions for bulk storage, but even the fastest floppy discs or bubble memories are much slower in operation than most logic families. The problem is not so much the rate of transfer of data, but rather the 'access time', i.e. the time taken between deciding that you wish to read or write to a specific location in memory, and actually being able to read or write that data. Although discs may be able to transfer data at many hundreds of thousands of bytes per second, it can take many tens of milliseconds to begin that data transfer. This is an unacceptably long time if it is to be repeated often within a system. As a result of this, extensive use is made of a much faster accessing memory technology.

SEMICONDUCTOR MEMORIES

Looking first at the RAM, its implementation is very straightforward since all that is required is an array of flipflops or latches! Any flip-flop or latch can be used to store a 1 or a 0, can be written into (by changing the logic state of the clock), and can be read from by simply looking at the logic state of the Q output. However, the design of typical CMOS flip-flops and latches is rather over-complex for a large memory system; too much space on the i.c. 'chip'



Fig. 8.1. The static memory cell

would be taken up with each device, and the total power consumption of the large numbers of devices involved would be prohibitive. As a result, a far simpler design of memory element or 'cell' is used, based on the 'bistable' circuit design. This is shown in Fig. 8.1, and is known as a 'static' RAM cell. ('Static' refers to the fact that data is stored in the cell for as long as power is applied; there is no decaying away of charge, or any other effect which may cause the cell to lose its data after a time.)

The memory is arranged in a matrix form; rows and columns of cells selected by control logic. To write a logic 1 into the static cell, the DATA column line for the column in question is taken to logic 1, and the DATA column line to logic 0. The word line, which determines the row in which the cell sits, is pulsed to logic 1, then back to 0 again. This forces the bistable into the state where point A is at logic 1 and point B at 0. (To write a logic 0, DATA column line = 1, DATA column line = 0, and again the word line is pulsed from 0 to 1 and back again). To read from the memory cell, the word line is pulsed again, but this time the logic states on the relevant column lines are monitored. If the DATA column line pulses to logic 1, then the bit stored was a logic 1. If the DATA column line pulses to logic 1, then the bit stored was a logic 0.

DYNAMIC MEMORY

The static memory cell contains four MOS transistors and two 'resistors' (also formed from MOS transistors), and hence is fairly bulky when used in quantity on an i.c. 'chip'. Supply current requirements are also relatively high, and this also restricts how closely packed the cells can be; too close together, and the resultant heat generated cannot be dissipated by the i.c. package, causing heat damage. Because of these restrictions, a lower power and more compact memory design is frequently used; the 'Dynamic' memory cell, as shown in Fig. 8.2.



Fig. 8.2. The dynamic memory cell

To write a logic 1 into the cell, the DATA column line is held at logic 1, and the word line is pulsed from logic 0 to 1 then back to 0 again. This stores a charge on the 'capacitive element', which is actually another MOS transistor since capacitors are very difficult to fabricate effectively on an i.c. chip. To write a logic 0, the process is repeated with the DATA column line at logic 0. To read from the cell, the word line is again pulsed and the logic state of the DATA column line is monitored; if it pulses to logic 1, the stored bit was also a 1, if it does not change at all then the stored bit was a 0. (The logic 1 pulse is caused by the 'logic 1 charge' on the capacitive element discharging, via the MOS transistor, into the circuitry monitoring the DATA column line).

There are two inherent weaknesses of this design. The first is that, unlike the static cell, reading is 'destructive'; any charge stored on the capacitive element is discharged into the control circuitry upon readout, so the memory has been lost. The second weakness is that the charge on the capacitive element leaks away after a very short time (the actual capacitance value is very small), so the memory is lost after a short time anyway—well under a second! To overcome both these problems, 'refresh' circuitry is used to rewrite any memory which has just been read with its 'own' data again, and to continually re-write all the data out of the memory and back into it again. These days, many popular dynamic RAMs have all the refresh circuitry built in, so they appear to the user to be static; no external refreshing is necessary.

THE ROM

Although the ROM is supposedly little more than a large array of pre-set switches, in practice it has a complexity similar to that of the RAM. This is necessary, to ensure that it can be addressed in a similar way to the RAM, that it can have a very high speed (i.e. the data should be fed out of the i.c. as soon as possible after the address has been selected), and in many cases that it can be programmed; it becomes a programmable read only memory, or 'PROM'. Naturally all ROMS have to be programmed at some time or other, so the term ROM is used for 'Mask Programmed' devices, i.e. those which have fixed programs built in during manufacture, and which are unalterable. 'PROM' refers to a device which is purchased 'blank', and is then written into; by applying highvoltage pulses into the device at the relevant address, the memory locations can be 'blown', rather like fuse links, to create the required pattern of O's and 1's. Because it is easy to make errors while programming in this way, most PROMS are made erasable; 'EPROMS'. Erasing such devices restores all the 'fuse links', enabling the programmer to start again from scratch, and is most usually done by exposing the i.c. chip, via a transparent window in the top of the i.c. package, to a specific wavelength of ultra violet light for typically between 10 minutes and an hour. This sets all the bits to a logic 1 state; when memory locations are subsequently 'blown', this sets the bits in question to logic O states. 'EEPROMS', or electrically erasable PROMs, have recently started becoming available, and use electrical erase signals; they become, in effect, a non-volatile semiconductor memory of fairly large size, although their erase and write times can be very long.

Non-volatile semiconductor stores are available in their own right, the Plessey 'NOVOL' range being a prime example. This is a range of counters, counters with 7-segment decoders, and one latch circuit, which have non-volatile circuit elements built in to retain the last logic states of the device for up to one year in the absence of power. In these devices, of course, only small numbers of bits are involved, so they are only a very small-scale memory in terms of ROM, RAM and bulk storage techniques.

MEMORY ORGANISATION

Most semiconductor memories are organised in terms of 1 bit, 4 bits, or 8 bits per address word. Hence, a memory device which claimed to be 256 words x 8 bits would have a total of 2048 cells within it, i.e. a total storage of 2048 bits. We refer to a unit of 1024 bits as '1K' bits (1024 being equal to 210;) hence, 2048 is 2K bits, 4096 is 4K bits etc. When memory devices are grouped together in a microprocessor or microcomputer system, we arrange them to suit the number of bits that the microprocessor uses. For example, the majority of microprocessors today work on an 8 bit system, so our memory will be arranged to be 8 bits 'wide' also. (By putting two 4-bit memories side by side, with common address lines, or eight 1-bit memories, etc.) Each set of eight bits is known as a 'BYTE', so a 4K Byte memory consists of 4 x 1024 x 8 bits i.e. 32,768 bits. Often, computer jargon shortens the term '4K bytes' to just '4K', which could cause confusion with 4K bits. It is usual to talk about individual memory i.c.s. in terms of bits, but assemblies of memory devices, memory systems, and microcomputers in terms of bytes. Typical 'personal' or 'home computers' are available with between 1K and 48K (bytes) of RAM, and up to 12 or 16K (bytes) of ROM.

THE MICROPROCESSOR

The 'micro', as it is known, is a very complex and technically advanced electronic device, and there certainly isn't room in this article to go into detail about its design and operation. However, we can take a look at the basics, to give an idea about the way that the technology is used, and the sort of facilities that can be provided. Two months ago, we looked at bus-orientated systems, and gave an example of some data manipulation in an A.L.U. using such a system. Let's now take those principles further, and look at a greatly simplified 8-bit microprocessor system. (Note that a microprocessor will often be referred to as a 'Central Processing Unit' or CPU.)

A block diagram of a simplified microprocessor is shown in Fig. 8.3. The internal 8-bit data bus, ALU, accumulator, and general purpose registers have all been met before in Part 6 of this series. The data and address bus controllers are basically bi-directional register and latch arrangements to permit the feeding of data and addresses to and from external switches, transducers, bulk memory, displays, etc. The address bus controller normally uses 16 bit words, which are made up from two separate 8 bit words obtained from the internal 8 bit data bus. Most of the general purpose registers are 8 bit, but there are some specific exceptions; the 'Program Counter' and 'Stack Pointer' registers are both 16 bit, for example.

The 'Instruction Register' is used, together with the Control Unit, to identify and verify specific binary numbers called 'Instruction Codes', which are fed into the microprocessor from the program written in the external ROM or RAM. These numbers provide specific instructions to the microprocessor; 'reset the accumulator' or 'increment' (add 1 to) register B, etc. When identified and decoded by the control unit, each instruction causes a specific number of control and timing signals to be sent out to the relevant registers and areas of circuitry, in order to cause that instruction to be implemented. The number of different instructions that can be performed by a micro is a very good indication of the 'power' of that micro, along with the number of registers that it makes available to the user. The popular Intel 8080A, for example, has 78 instructions in its instruction set, whereas the Zilog Z80 (acknowledged as one of the most powerful 8 bit microprocessors available) has 158 instructions, including all 78 of the 8080. The Z80 is also well sup-



Fig. 8.3. Simplified 8-Bit microprocessor block diagram

ported with registers; a total of 18 8-bit and 4 16-bit registers are provided.

Unlike a conventionally designed piece of circuitry, the microprocessor based system is incapable of functioning, even when mechanically complete and apparently finished. The mechanics—p.c.b., power supply, i.c.s resistors, etc., are collectively known as 'hardware'. What is needed is a 'program'; a set of binary numbers which instruct the microprocessor what to do, and how to do it. Such programs or lists are known collectively as 'software', they are, in effect, the paperwork involved in the system. Programs, of course, can be written into ROM, RAM, or bulk storage. In this case, these devices consist of both hardware and software, and so any memory device containing a program is known as 'Firmware'; half way between being hard and soft!

PROGRAMMING

A program contains a series of instructions, and often data too, although in many systems the data is provided externally by switches, transducers, etc. The microprocessor executes each instruction in sequence, one at a time, although there can be built-in jumps such that upon receipt of a specific instruction the microprocessor will 'move' to a different area of the program, before or after the point which it has just 'left' and continue executing the program from there. The sequence of instructions given at the end of Part 6 of this series, covering the addition of two numbers together and the putting of the result in register 'D', is an example of part of a very simple program.

Programs can be shown diagramatically by a 'flow chart'; a set of symbols and text, which is the software equivalent of a block diagram used to represent the hardware. Let us take, as an example, the requirement to flash an l.e.d. on and off for a certain number of times as determined by a thumbwheel switch. The simplified flow chart for this program is shown in Fig. 8.4. The 'loop' is gone round as many times as the l.e.d. is flashed on. The type of instruction causing the program to jump from the question 'is the content of register B zero?'', to the ''turn on l.e.d.'' sequence is known simply as a 'jump' instruction. In fact, the probable instruction would be ''jump if not zero'', which is an example of a 'conditional' jump. These conditional jumps make programming very much easier, because they effectively test for a specified condition (''accumulator = zero'', or ''a certain logic state = 1'', etc.,) then either jump to a different part of the program, or continue to the next program step in the list, depending on the result.

The turning on of the l.e.d. would be accomplished by an external circuit driven from the microprocessor's 8 bit data bus, and the thumbwheel input would likewise be fed into the micro. The half second delay could be created by using an external monostable circuit, again, via the data bus, or by software methods; the computer takes a finite time to perform each program step, so by making it do a 'trivial' exercise a few tens of thousand of times, then time can be wasted, and a suitable delay created. In all programs, the most important thing to remember is that within the circuitry in question, both the instructions and the data are being represented as logic 0's and 1's; simple binary numbers.

THE USE OF MICROPROCESSORS

There are, essentially, two ways of using a 'micro' within a system. The first is 'dedicated' usage; the microprocessor has a single fixed program held in ROM, which it always per-



EG779

Fig. 8.4. Simplified flow chart of I.e.d. flasher

forms in the same way, while using small amounts of RAM for temporary storage if necessary. The user will have no means of re-programming this device. The microprocessor in this application is simply replacing large amounts of conventional logic circuitry. Test instruments, hand held electronic games, sewing machines, washing machines, cassette recorders; etc., whenever they use a microprocessor, they do so on a dedicated basis, and in each case the micro gives a fixed set of responses to a fixed set of stimuli, or inputs.

'Non-dedicated' use introduces us to the domain of the microcomputer, where the full performance, instructions, and capabilities of the microprocessor are fully at our disposal.

THE MICROCOMPUTER

The term 'microcomputer' means more than just a small computer. It is a computer in which the CPU is manufactured as a single i.c.; a microprocessor. Unlike the dedicated micro-based systems, it requires definite operator interaction before it can be of very much use at all! It is a programmable device, so without the program written into it, it can do very little. The constituent parts diagram of a typical microcomputer is shown in Fig. 8.5.

The video output of the microcomputer is normally in the form of a 9 to 12 inch CRT; often a portable domestic television set. This video monitor is usually arranged to display a number of ASCII characters corresponding to the binary numbers held in a particular part of the memory. Imagine the



Fig. 8.5. Constituent parts diagram of a microcomputer system

screen being divided up into a matrix of squares, with each square having a specific and unique address. There are typically around one thousand (1K) of these. Part of the RAM known as the 'Video RAM' is usually dedicated for this use, and the display is then known as 'memory mapped'. The keyboard enters ASCII codes into the computer, which are normally fed up onto the screen as the program is being written. The bulk storage, printer, and external outputs and inputs are all fairly self-explanatory.

The microprocessor has many extra circuits and devices surrounding it; encoders and multiplexers to feed the keyboard information in, character generators and modulators to enable the memory to be displayed on the screen, and (of course) ROM and RAM. The ROM in a microcomputer normally carries the 'monitor program' (not to be confused with the video monitor); this is the 'apparent intelligence' of the computer, and is the interface point between the microprocessor logic, and the human operator and keyboard. Depending on the size and complexity of this monitor program, the facilities offered to the user can be considerable. The monitor contains basic and fundamental information within the computer, to tell it to scan the keyboard for any keys pressed, then to feed the ASCII character corresponding to the key pressed into the section of video RAM corresponding to the relevant position on the screen. It also enables, editing of the video RAM; deletion of characters, moving the display up and down ('scrolling'), etc.

Although the microprocessor operates on binary numbers, known as 'machine code', the human operator can program the computer in the hexadecimal equivalent to those binary numbers, making it quicker to type in, and easier to understand. This hexadecimal listing is known as 'object code'. Some systems help still further, by allowing the program to be written in 'mnemonics', which is a means of writing hex instructions in an abbreviated English language format. For example, 'JP' could be used to represent a jump instruction, the hex code for which might be C3, or 'INC B' might be used to represent the instruction 'increment the contents of register B', who's hex code might be 04. By using mnemonics (pronounced 'nimonics'), the program (when written down or typed in) can be much easier to follow and understand; the instructions all mean something and are no longer a collection of fairly arbitrary looking letters and numbers!

To produce machine code from the mnemonic program, an 'assembler' program must be used, which acts as a 'converter' between mnemonic and binary. (To translate back from binary to mnemonic, a program called a 'disassembler' is brought into use.) Much of the programming for dedicated microprocessors is written on microcomputers in object code or mnemonics, because this level of computer operation is very fast; the computer operates very efficiently when using these codes. However, programming can still be a complex, slow, and difficult task, as the detailed internal operation of the microprocessor is still dominant; programs are written in terms of registers, clock signals, ALUs, flags, etc. To bring the program one step higher in level, to a nearer approximation to the English Language, we must use a 'high level language'.

HIGH LEVEL LANGUAGES

The high level language raises programming to a fairly easily understood level. Instruction such as 'GO TO', 'PRINT', and 'IF $X = Y \dots$ ' are used, and enable people who are not immediately familiar with the computer or program in use to make sense of it all fairly quickly. There are many different languages; FORTRAN, COBOL, ALGOL, BASIC, PASCAL, FORTH, etc., the most commonly used of these (in the case of microcomputers) being BASIC, which stands for 'Beginners Allpurpose Symbolic Instruction Code'. Although there are several 'dialects' of BASIC they all share many common properties, and BASIC programs for one make of computer can usually be run on other computers with only a small amount of modification, if any at all. The opposite is true of object code and mnemonics, since these are specific to the microprocessor and monitor program in use, and can be quite difficult to translate for other systems. A special program must be used within the computer to convert programs, written by the user in BASIC, into machine code, and to convert machine code 'replies' from the computer back into the relevant BASIC statements and comments. This special program is normally held in a ROM, as firmware.

High level languages, because of the extra complexity involved, are relatively slow in use; each step of the program takes longer to execute than in the case of machine code based programs. Even so, the operation can be fast enough for all but the most demanding applications.

GETTING TO GRIPS WITH THE 'MICRO'

Unfortunately, the microprocessor is not something that you can just 'dabble' at; you can read all that you like about it, but until you get 'hands on' experience with a micro-based system you will still know very little about the practicalities of the subject. By far and away the best learning aid is a microcomputer itself, which will cost a significant amount of money, so it is important to shop wisely. There are microcomputers on the market that may seem attractive because of their very low cost, and some good publicity. Indeed, they demonstrate considerable engineering achievement in squeezing so much circuitry into remarkably small spaces, for a reasonable price. However, they can be very restricting in performance; the keen user will find that the capabilities of the machine, even with 'expansion' (more memory), can be very limiting. For what they are, they are excellent, but they are not fully fledged microcomputers in the same class as machines costing over £200. It is strongly suggested that if you are serious about learning about microprocessing and computing, you spend more money to avoid disappointment in the very near future. As a compromise, some computers in kit form can enable you to start off with a simple and inexpensive system (working on the object code level), then build it up slowly, and add a high level language, to finish up with a system which is worth several hundred pounds.

The machine to buy is very much a personal choice, to be made in combination with a study of magazine reviews, advice gleaned from other, more experienced microcomputer users, and a trip or two to your local dealer. Obviously, you get what you pay for, but to a certain extent you can pick and choose your approach. For those interested in the hardware side of the subject, the kit computers mentioned earlier can be a good starting point, but only if your standard of construction, especially soldering, is very good. Kit computers are NOT for beginners! Nascom 1, Nascom 2 and Tangerine products, amongst others, can certainly be recommended. Ready-built computers are for those less confident in the hardware side; Tangerine, Video Genie, Sharp and Tandy all have ready-built products, and the new 'BBC micro' is attracting a lot of attention at the moment, although at the time of writing it is not yet on the market. Paying out still more can provide small business computers such as the PET, APPLE or Tandy, but these represent a considerable investment and are not necessary unless you are likely to need their very advanced facilities later. A choice in the microprocessors used comes down to 6502 versus Z80 in the case of most microcomputers mentioned. The Z80 is undoubtedly more powerful, but the 6502 is somewhat more widely used, so once again the choice is a personal one.

Another point to make is in the interest of domestic harmony! Computing can take up a surprising amount of time. 'De-bugging' programs has kept most people busy into the early hours on a number of occasions, and you should be aware that demands on your time can be considerable, especially at first. Computing is not something that you can do a little at a time; several hours at a stretch is needed to get to grips with the problems, and preferably in peace and quiet, too. Don't let this put you off, though—it's a useful and fascinating subject, and well worth spending time and effort on!

In conclusion, if microcomputing or microprocessors are of interest to you, then the only real way to learn is to buy a microcomputer, and spend some time with it. Before committing yourself, why not see if the local school or technical college has evening classes in computing or microprocessors, or even runs a suitable club? This would enable you to try out various machines before buying, learn more about the subject, and ask the advice of other enthusiasts.

IC TECHNOLOGY

As technology has advanced through the years, so the number of active devices on an i.c. chip has increased. We can define the 'scale' of integration as follows (remember that each gate is made up of a number of transistors or MOS devices):

Small Scale Integration (SSI); up to 12 gates per i.c. Most NAND, NOR, inverter, and similar combinational logic i.c.s. fall into this category.

Medium Scale Integration (MSI); from 12 to 100 gates per i.c. This covers the range of counters, registers, decoders, flip-flops, etc.

Large Scale Integration (LSI); from 100 to several hundred gates per i.c. Multiplexed counter/driver i.c.s. digital clock i.c.s, and similar complex devices all fall into this category.

Very large Scale Integration (VLSI); from several hundred to several thousand gates per i.c. This is the realm of the microprocessor and the large memory i.c.s. Some micros even have memory i.c.s. both ROM and RAM, built in! Many 'accessory' i.c.s to the microprocessor are needed, such as programmable timers, peripheral management devices, memory controllers, input/output controllers, etc., and most of these are LSI or VLSI devices in their own right. In many cases, the microprocessor is only a very small part of the whole system; its support i.c.s can take up a lot more space and cost more than it does itself!

To help cut down on the number of 'accessory' i.c.s in any logic circuit or system, the 'Uncommitted Logic Array', or 'ULA' has been developed. This is a very large collection of logic gates which have no pre-determined interconnections. By designing a suitable pattern of interconnections of these gates, a custom-built i.c. can be created which will do the job of a large number of 'off the shelf' devices. Because the ULA is a standard device, with only the final interconnection layer being customised for each application, it is considerably cheaper than having an entire i.c. designed and built, and this approach, therefore, holds great promise for further miniaturising circuit designs in the years to come.

DIGITAL ELECTRONICS IN THE FUTURE

The pace of technological advance is so fast at the moment that it is difficult, if not impossible, to predict developments in the medium or long term future. However, it is possible to extrapolate current trends and developments, and to fairly accurately assess the changes that we can expect to see in the next few years.

In terms of microprocessors, the trend is towards larger and larger numbers of bits being used in a device. The industry 'standard' is currently 8 bits. 16 bit micros are now readily available and are starting to be used in large quantities, yet already we have seen the first appearances of the 32 bit micro; an incredibly powerful i.c. which will enable us to out-perform the much larger 'mini-computers' with single board VLSI microcomputer systems. 64 bits is still long way off, but is far from being an unrealistic or impossible dream. To keep pace with the speed and power of these large devices, memory manufacturers will continue increasing the number of cells per chip, and increasing their operating speeds, probably with dynamic RAM, EEPROMS, and bubble memories showing the greatest improvements.

The 'in' subject at the moment is voice synthesis and speech recognition systems. Cheap, high quality synthesis i.c.s are readily available at low cost but have the severe restriction of a very limited vocabulary. Infinitely variable 'phonetic' synthesisers are more expensive, and most of them suffer from a very poor quality of speech. However, the i.c. manufacturers and designers seem determined to thrust the 'talking hot drink vending machine', and similar ideas, upon us, so this is sure to be an area of considerable future expansion! Speech recognition is a much more difficult task, due to the subtle nuances, inferences, dialects and idiosyncrasies that punctuate the speech of all of us. Again, systems are available with limited performance, and these will be improved upon considerably in the years to come.

Combining speech synthesis and recognition will allow direct communication between the computer and the 'manin-the-street', which offers vast scope for technological advance and integration of the computer into our everyday life. From the other angle, the man-in-the-street is becoming more familiar with the subject of computing in its own right. All British schools are being equipped with at least one microcomputer, and already many schoolchildren are achieving stunning levels of proficiency and knowledge about computing. The 'electronic office' is with us; word processing, facsimilie machines, electronic filing, and data transfer systems are all available and are being used in many of the large companies. Expensive, yes, at the moment, but as with most electronic goods a continued demand will result in dramatically lowered costs.

Returning to technological advances at the device level, the Uncommitted Logic Array, as mentioned earlier, will help in the miniaturisation of circuit assemblies, which in turn will increase the number of everyday products which have electronics incorporated into them. Speed limitation is a major problem in many microprocessors and computers, and advances are already being made towards providing the technological basis for the next generation of devices. Gallium Arsenide semiconductors, and those using the 'Josephson Junction', both of which need very low temperature cooling systems to operate, offer considerable decreases in gate propagation delays, with the resultant increase in computing speeds, and show great potential for the future. The one thing that seems certain about the times ahead is that they will be filled with change, challenge, and interest, just as the last few years have been. The profession, and hobby, of electronic engineering is guaranteed to have a colourful future for many years to come!

DIGITAL DESIGN TECHNIQUES

We now conclude this series on digital design techniques; there's no 'mini-project' this month because of the level of complexity that we have now reached, and the resultant limitations of space, so for the final project see next month's issue. In the past few months we have worked through from binary logic and combinational gates, all the way up to microprocessors. The pace has been swift, and to get full benefit from the series it is strongly suggested that you follow up the points made, and do further reading, experimenting, and designing of your own. There is much to be learnt and a great deal of fun and enjoyment to be had, so start putting together your own projects today!

ACKNOWLEDGEMENT

I would like to express my grateful thanks to my father, Peter Gaskell, and to my friend and colleague John Miller, for all their help, criticism, and suggestions in the preparation of this series. T.A.G.



Our book *PE Popular Projects* containing a selection of popular projects is now available. The book costs £1.25 from retail outlets and is also available for £1.50, UK post paid, or £1.80 overseas surface post paid, from Post Sales Department (PE Popular Projects), IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 OPF.

The book contains the following projects:

Battery Voltage Indicator, Rev. Counter, Ammeter, Engine Temperature Meter, Dwell Meter, Headlight Warning, Automatic Car Aerial, Digital Temperature Controller, Ultrasonic Burglar Alarm, Home Freezer Alarm, PE Diamatic, Digital Exposure Timer, Smooth Fuzz, Phaser, Guitar Sound Multiprocessor, R.C. Failsafe, Waveform Generator and Pulse Generator.



PART 5

and the second second second

RICHARD BECKER · TIM ORR RICHARD MONKHOUSE

BOTS



RESUME

Genesis is a series of hydraulic computer controlled robots ideal for education, light industry and as microcomputer peripherals.

Features are: extremely low cost, up to six controllable axes, positional sensing, servo control, continuous path motion, learning ability, RS232 interface, and IR remote control for the mobile unit.



Two photographs are shown here; one of the Genesis Mobile unit, M101, and the lower photograph which illustrates through multiple exposure photography, the level of robot technology with us today, by way of the Cincinnati Milacron machine. This latter machine (photo courtesy of Cincinnati) is in a somewhat prohibitive price range for the student. *Education* is quoted as an ideal application for the Genesis robots because they are *inexpensive* reprogrammable multi-tasking machines: and *light industry* because they are also engineered to a professional standard.

Constructor's Note

Complete kit of parts for this project can be obtained from Powertran Cybernetics, Portway Industrial Estate, Andover, Hants SP10 3WN. & Andover (0264) 64455.

Prices are as follows

Genesis M101 4 axis model (excluding wheel base)	£295.00
Genesis M101 5 axis model (excluding wheel base)	£345.00
Genesis M101 wheel base	£79.00
Genesis P101 4 axis model	£450.00
Genesis P101 6 axis model	£545.00
Genesis S101 4 axis model	£355.00
Genesis S101 5 axis model	£405.00
Position detector coil set for M101, S101 4 axis models	£15.00
Position detector coil set for M101, S101 5 axis models	£19.00
Position detector coil set for P101 4 axis model	£15.00
Position detector coil set for P101 6 axis model	£24.00
Position detector board for M101, S101 4, 5 axis mode	ls £6.50
Position detector board for P101 4, 6 axis models	£7.50
Motor drive board for M101 wheel base	
(2 required per machine)	£11.50
Control electronics for M101 (microprocessor board,	
interface board, display board and mounting	
bracket)	£135.00
Processor box for S101, P101 (microprocessor board,	
interface board, display board, power supply,	
interface cables, conduit, cabinet)	£175.00
Parts for RS232C interface (fits on microprocessor	
board)	£14.50
Hand held controller box for M101 (includes	
infra red transmitter and rechargeable battery)	£47.00
Hund held controller box for S101	£33.00
Hand held controller box for P101	£33.50
All prices subject to 15% V.A.T.	

Components . . .

CONTROL BOX (MOBILE) Resistors

R1, R4, R6 R2 R3, R5, R23, R29 **R7** R9, R12, R14, R15, R28 R10, R11 R13 R16 R17 R18, R22, R24 R21, R27 R19, R20, R25, R26 All resistors 1W 5%

180k (3 off) 120k 47k (4 off) 470 3k9 10k (5 off) 1M (2 off) 2k2 33 100k (3 off) 560k (2 off)

12k (4 off)

Potentiometers

VR1	100k slider (linear)
VR2	47k slider & centre indent (linear)
VR3	470k preset
VR4, VR5	1M preset (2 off)
VR6, VR7	100k preset (2 off)

Capacitors

C1	1n Siemens B32560	
C2	10n ceramic Piher	
C3, C4	39p ceramic Piher (2 off)	
C5, C6	100µ/16V elect. Axial lead (2	2 off)
C7, C8	10n Siemens B32560 (2 off)

Transistors and Diodes

TR1 TR2 TR3, TR4, TR5 D1-6, D9-14, D17, D18, D21, D25-30, D33-38, D43 D41. D42 D44, D45

BC184L TIP31A BC182L (3 off) 1N4148 (28 off) TIL 38 (2 off)

9491BJ (1V2 Band Gap) or LM385 (2 off)

Integrated Circuits

IC1, IC2	4051B (2 off)
IC3	4024B
IC4, IC9	4013B (2 off)
IC5	40106B
1C6	4023B
IC7	4011B
IC8	LM324
IC10	ICM7555 IPA
IC11	10k s.i.p.

Miscellaneous

Printed circuit board (RTX) M101 metal cabinet and fixing screws PP3 stud connector PP3 rechargeable battery Slider knob Pot knob 8-pin d,i.l. socket 14-pin d.i.l. socket (7 off) 16-pin d.i.l. socket (2 off) Switches (27 off) Slide switch 3.5 jack socket Jack-to-jack lead (1 metre)

COMPONENTS

CONTROL BOX (P101 & S101)

Resistors

R1, R4, R6	
R2	
R3, R5	
R7	
R10, R11	
R12	
R13	
All resistors 1W 5	%

180k (3 off) 120k 47k (2 off) 470 1M (2 off) 10k 2k2

Capacitors

C1 C2 C3, C4 **C**5

1n Siemens B32560 10n ceramic Piher 39p ceramic Piher 100µ/16V axial lead

Transistors and Diodes

D1-6, D9-14, D18, D21, D25-30, D33-38, D43 D46 TR3

1N4148 (27 off) 5V6 Zener BC182L

IPA

Integrated Circuits

IC1, IC2	4051B
103	4024B
IC4	4013B
IC5	40106B
IC6	4023B
IC7	4011B
IC10	ICM7555
IC11	10k SIP

Miscellaneous

Metal cabinet and fixing screws Printed circuit board DIN plug 5-pin 240° 5-way cable (3 metres) Slide switch push-button switches 8-pin d.i.l. socket 14-pin d.i.l. socket (5 off) 16-pin d.i,l. socket (2 off)

Note: In part 2, Interface board components list, under Miscellaneous-SIP1 and SIP2 should be considered as IC22 and IC23. SIP3 is IC24.



Fig. 1. Robot controller *RS232 version only. **Not used at present

INTERFACE BOARD

The Robot requires many more than the 20 I/O lines available from one PIA. To obtain these without going to multiple PIA's, the interface board uses a secondary 8 bit bus connected to one side of the PIA whilst the other side generates latch clocks for output latches or output enables for system inputs. Although careful attention to detail was necessary in the software to avoid bus clashes, this method makes efficient use of board space and means that the software can be debugged easily from an external computer.

The outputs from the expanded I/O scheme comprise two sets of 8 bits for controlling the solenoids and the motors, 8 bits for the display and bleeper and 8 bits for extra devices. On the input side there are 2 bits for the keyboard decoder interface one bit to signify RS232 connection and 5 bits for extra inputs.

SYSTEM DESCRIPTION

On manual control which is available in the EDIT mode, the robot arms can be moved around using the appropriate control buttons on the control box. There is logic to prevent both solenoids for an axis being activated simultaneously. The robot is taught by being moved under manual control to the desired position. Pressing the INSERT button enters the position and the microprocessor stores it in the battery backed-up CMOS RAM.

When a position is to be remembered, a combination of an analogue multiplexer and an eight bit analogue-to-digital converter (interface board — IC25 and IC24) selects each of the 5 transducer voltages in turn, and digitises them.

On replay, the ADC/multiplexer combination is constantly

scanning the 5 inputs under software control, the measured positions of the five axes being compared by the microprocessor with the current target positions. These comparisons are used to activate the correct solenoids to achieve the target positions.

Manual control of the drive motors (on the mobile unit) is made variable by mark/space modulation of the motor drive switching transistors. These mark/space modulation timings are derived by the program from an internal look-up table that gives the mark/space ratios for the left and right motors from the position of the panel Speed slider and the Steering pot. These timings are themselves also generated by the microprocessor.

RS232 input

To control the robot from an external computer, an RS232 input is provided. If the RS232 lead is plugged in when the unit is powered up, the unit will ignore the keyboard and wait for a command from an external computer.

A move command is simply sent as a string of ASCII characters. The format consists of an "M" followed by six pairs of hexadecimal digits (each pair representing a number between Ø and 255 — the desired position for each axis). The last pair is the jaw command, and should beØØ for "jaws open", and FF for "jaws closed", e.g. M8Ø8Ø8Ø8Ø8Ø8ØFF will move all axes to a central position and close the jaws.

When the move has been completed the robot will send back a "*" down the RS232 line to inform this to the sending computer.

On five axis units, the format is the same, but the last-butone pair of digits is ignored.



MICRO-EUS

THE TWO topics in this month's Micro-Bus show how micros can be used in amateur radio. The first program receives fax transmissions, and displays them on a computer screen. The second program acts as a Morse-code tutor, for those trying to learn Morse code. Both programs were developed by Steve Platt, a radio amateur in Cambridge (G6AZ1).

RECEIVING FACSIMILE PICTURES

Pictures can be transmitted over channels normally used for speech, such as the telephone, or over the air by radio, using a system known as 'facsimile' or 'fax'. This system is used by radio amateurs to send diagrams such as weather maps and charts on the HF and VHF bands, but one problem faced by amateurs wishing to try their hand at fax is displaying the received pictures without expensive equipment.

The program to be described solves the problem by using the high-resolution fourcolour graphics on an Acorn Atom computer to display fax transmissions on a TV screen straight off the air from a receiver. It was developed by *Steve Platt* of Cambridge (G6AZI), and what follows is based on his description.

"Radio amateurs generally use fax on the HF bands (short waves) where many commercial transmissions can be received. Fax is also used on the popular VHF 'two-metre' band where the calling frequency is 144.700MHz. A sample fax picture, of a street scene, is shown in Fig. 6; it is composed of three brightness levels, and has a total resolution of 128 \times 192. It was received on 144.675MHz FM using the program of Fig. 7 with the receiver output connected directly to the Atom's cassette input!



Fig. 1. Facsimile picture decoded and displayed by the program of Fig. 2

10 DIM LL(18), JJ(3), B(100) 20 DIM D(100) 20 DIM DI100) 30 FOR N=0TO18; LLN=TOP; NEXT 40 FOR N=0TO3; JJN=TOP; NEXT 105 PRINT \$12, "FAX" 110 PRINT *ENTER COUNT AND SYNC* D 115 PRINT '"TRY 77 & 8"' 120 INPUT C.S. GOSUB 2000 225 GOSUB 2000 130 FOR N=0 TO S; B?N=2; NEXT 140 B?(S+1)=0 150 FOR N=5+2 TO 100; B?N=3; NEXT 200 REM DO IT 1 210 CLEAR4; COLOUR 1 220 DO 220 DO 225 DO UNTIL?#B001<>#FF 2084=0; 2085=080; LINKJJ3 UNTIL 0 230 240 2000 REM - COUNT CYCLES 2004 PRINT \$21 2005 FOR N=1 TO 2 2007 DIM P(-1) 2010 [2020: LL1 LDA @0 2020:LL1 LDA @0 2030 STA #82; LDA @C; STA #80 2035 LDA @1; STA #81 2040 LDA @@20 2050 BIT #8002 2060 BEQ LL2 2070:LL4 DEC #80 2080 BEQ LL9 2080 BEQ LL9 2090 NOP 2100 JMP LL10 2110:LL10 JMP LL11 2120:LL9 DEC 081 2130 BEQ LL3 2140:LL11 LDA 0020 2150 BIT #8002 2160 BEQ LL6 2170 JMP LL15 2170 JMP LLIS 2180:LLIS JMP LLS 2190:LLS JMP LL4 2200:LL6 INC 082 2210 JMP LL2 2220:LL2 DEC 080 2230 BEQ LL12 2240 NOP 2250 JMP LL13 2260:LL13 JMP LL14 2270:LL12 DEC 081 2280 BEQ LL3 2290:LL14 LDA @#20 2300 BIT #B002 2300 BIT @B002 2310 BNE LL7 2320 JMP LL16 2330:LL16 JMP LL8 2340:LL8 JMP LL2 2350:LL7 INC %82 2360 JMP LL4 2370:LL3 LDA @82 2380 RTS 2400:LL17 LDX @0 2410:LL18 LDA @C; 2400:LL17 LDX 00 2410:LL18 LDA 0C;STA 080 2420 LDA 01; STA 081 2430 JSR LL1 2440 STA D,X; INX 2450 CPX 0101 2460 BNE LL18 2460 BNE LL18 2470 RTS 2500:JJ3 JSRLL1; CMP@S; BCC JJ3 2500:JJ2 JSRLL1; CMP@S; BCS JJ2 2510:JJ0 LDX @4 2520:JJ1 JSR LL1; TAY; LDA B,Y 2530 RORA; ROL@83; RORA; ROL@83 2540 DEX; BNEJJ1 2550 LDA@83; STA(@84,X); INC@84 2560 LDA@84; AND@31; BNEJJ0 2570 LDA@84; BNE JJ3; INC@85 2580 LDA@85; CMP@@98; BNE JJ3 2590 RTS 2960]` 2470 RTS 2960] 2970 NEXT N 2980 PRINT \$6 2990 RETURN 3000 END

Fig. 2. Program for the Atom displays facsimile pictures

Compiled by DJD.

"The picture was transmitted (by G8RYL) using a surplus fax machine fed into the transmitter microphone socket. The picture is wrapped around a rotating drum, and scanned with a photocell. The photocell moves slowly along the drum to give a vertical scan, while the drum's rotation provides the horizontal scan. The output from the photocell is then used to vary the frequency of an oscillator with the brightness of the picture. Drum speeds of 60, 90, 120 and 240rpm are most popular; the program to be described works with 120rpm. The picture's "aspect ratio" is varied by setting the number of drum rotations for each inch that the photocell moves. Further details may be found in Radio Communication, August 1978, and Wireless World, December 1976 and March 1977.

PROGRAM DESCRIPTION

"The main task of the program is to determine the audio frequency of the incoming fax signal. This is done by counting "zerocrossings" during fixed intervals of time. The number of zero crossings is then related to a scale of brightness levels, and a point of the corresponding brightness is put on to the Atom screen by poking directly into the graphics memory.

"First the vectors and labels are declared, and the labels are cleared (lines 10 to 40). Variables C and S are parameters which can be varied to tune the progam to the machine being received; the values given, C = 77 and S = 8, seem to work best. The time interval over which zero crossings are counted is determined by C, and S determines the number of zero crossings that correspond to the fax syncpulse level; i.e. the lightest part of the picture. The call to subroutine 2000 in line 125 assembles the machine code; note that this code uses C and S as constants, so these must be set before assembly.

"Vector B is a look-up table of brightness values. If X zero-crossings are counted, the colour of the point plotted will be B?X. Values of 2, 0 and 3 correspond to white, grey and black respectively. The brightness levels are set up in lines 130 to 150, based on the value of S selected.

"The main program loop, in lines 210 to 240, first waits for a key-press; locations # 84 and # 85 are then pointed to the start of the screen memory (line 230), and the machine-code routine JJ3 fills the screen.

ASSEMBLER ROUTINE

"Lines 2000 to 3000 constitute the assembler routines to receive a fax transmission, and display it in three brightness levels on an Atom screen. Routine LL1, from 2020 to 2380, is the key subroutine and is actually

much simpler than it looks; it acts as a software frequency meter, counting the number of zero crossings on the cassette input in a fixed time interval.

"Each time the main loop is executed it reads the cassette input, and increments location # 82 if the input has changed. The main loop takes 32 usec, and is balanced with NOPs so that each decision branch takes the same number of cycles. The number of times the loop is executed is determined by locations # 80 and # 81; these are set to C in lines 2030 and 2035, and although in the present application C was only 77, a two-byte counter was used to allow expansion to longer measurement times. With this value of C the routine counts zero crossings for about 2.5 msec, before returning with number of zero crossings in location # 82.

"As a debugging aid, lines 2400 to 2470 will dump 100 samples of the input frequency in the vector D. Typing:

LINK LL17

followed by printing the values D?0, D?1... D?100 will allow the input to be checked. If the values are not the range 7 to 12 the value of S may need to be changed.

DISPLAY ROUTINE

"The display routine, JJ3 in lines 2500 to 2590, is called from the main BASIC loop. It waits for the synchronising pulse which precedes each line of video information (lines 2500 and 2505), and then fills a line on the screen. To do this it calls the LL1 routine to sample the input frequency, then looks up the corresponding brightness value in vector B (line 2520). This is a 2-bit pixel value which is shifted into the screen memory via a work byte, # 83, in line 2530. When four pixels have been packed the byte is updated in the graphics area (line 2550). At the end of a line (when the address is a multiple of 32-line 2560) the program goes back to wait for another sync pulse.

"The remaining part of the routine updates the upper byte of the screen address, and checks for reaching the bottom of the screen, address # 9800 (line 2580).

POSSIBLE EXTENSIONS

"The facsimile system described above is similar to the slow-scan television (SSTV) system used by amateurs, employing modified TV cameras, and the same program has been used, with slight modification, to receive SSTV "conversations" on the HF bands. Since the sync levels are reversed, lines 2500 and 2505 should be exchanged, and the values of C and S may need to be altered.

"Thought is also being given to an Atombased fax transmission system built around some sort of art/graphics sketchpad program. This would allow the user to sketch a pictorial reply to the received image on the Atom screen, using keyboard input.

MORSE-CODE TUTOR

Many radio amateurs are put off from using the short-wave bands because of the need to learn Morse code to obtain a licence, and the main problem is finding a source of slow Morse to learn from. The program for the Acorn Atom, shown in Fig. 3, acts as a Morse tutor by producing a continuous stream of letters in Morse code. The Tutor program lets you learn a subset of the alphabet, starting with the letters which have the shorter codes. In addition, the speed and spacing of the Morse produced can be varied to cater for different abilities. The program presents letters in groups of five, separated by a single space. This is similar to the test format used in the amateur radio examination, and so the program can be used to test whether the required standard has been reached. The program can also be used to learn letters, by watching the words on the screen as the Morse is played.

Many people have their own theories as to the best way of learning Morse code. Generally it is thought to be better to hear the letters at full speed from the beginning. This way you identify the sound of each letter without thinking of a written pattern of dots and dashes, which can be a distraction when trying to gain speed! It is therefore best to start with the speed set to your target speed, and have extra time between letters, over and above the normal spacing. The Post Office test for radio amateurs requires a speed of 12 words per minute, where words are defined as having five letters; other tests may need less.

CODE STORAGE

The code for each letter occupies two bytes, in vectors C and L. The byte in C holds a binary sequence which gives the code for that letter, with '0' for a dot (short bleep) and '1' for a dash (long bleep), starting with the lowest bit. The byte in L gives the total length of the code; i.e. how many bits to use from C. As an example, take the letter Z, which is letter 26. The value of L?26 is 4, so taking the bottom 4 bits of C?26 we have 0011. Thus the Morse code for Z is:

Only the codes for the letters are stored, although the program could easily be extended to numerals and punctuation.

PROGRAM DESCRIPTION

Lines 10 to 25 allocate space for vectors, C, L, S, and A, for labels VV0 to VV8, and for the machine code at P. The values of C and L for the Morse codes of the letters A to Z are set up in lines 100 to 170, and the assembly code is then assembled (line 180). The string S contains the alphabet in the order of shortest letters first; i.e. in rough order of difficulty. The program presents subsets of the alphabet starting with the first letter, S?1.

The main control loop of the program, from line 200 to line 295, determines how the codes are presented, and this could be altered to suit individual requirements. For example, the program could present codes one letter at a time, with you typing in your guess. K determines how many different letters from string S are being taught; thus, with K = 3, only letters E, T, and A are presented. G can be set to a value greater than 0 to give extra spaces between letters when learning (line 205). The program then presents a total of 24 'words', each of which is a 5-letter group (lines 210 and 220) picked at random from the first K letters in string S (lines 240 and 250). Each letter is 'bleeped' in Morse code by subroutine 'c' (line 260), followed by any extra spaces wanted between letters (line 265). After each word

there is an extra space (line 275), and the word is printed to the screen (line 280).

BLEEP ROUTINE

The assembler code in lines 900 to 995 of Fig. 3 is a routine to bleep the Atom's speaker, and is similar to the one in the Atom operating system. Location # 80 gives the pitch of the tone and location # 81 the length. The same routine is used for dots, dashes, and gaps, with location # 82 set to 4 for a tone and to 0 for silence; this ensures that tones and gaps are all exactly the right length.

An additional assembler routine at VV3 (line 982) gets the ASCII value of the next key pressed in location # 83.

```
5 PRINT $12, "WORSE TUTOR"'
10 DIM C(26),L(26),S(27)
25 DIM A(5),V(8),P(-1)
10 REM SETUF TABLES
110 C(0-#$50L0200;L(0-#4040200
120 C(1-#$304001;L(1-#3030401)
130 C(1-#$50L0000;L(1-#3030402)
140 C(12-#7010302,L(12-#30202)
150 C(1-#7010302,L(12-#303040)
160 C(120-#608040),L(120-#30402)
170 C(12-#30D00;L(12-#30402)
170 C(12-#30D00;L(12-#30404)
170 C(12-#30D00;L(12-#30402)
170 C(12-#3000)
170 C
```

Fig. 3. Morse-code tutor for the Atom

MORSE CODE OUTPUT

The final routine, 'c', outputs the code for the letter in I, where I = I gives A, up to I = 26 for Z; any value outside this range is treated as a space (inter-letter gap). The routine uses vectors C and L, set up as described above. Each dot is a bleep of length T (lines 1025-1040), and each dash a bleep of length 3*T (lines 1050-1052). There is a gap of length T between each dot or dash (lines 1065-1070), and a gap of length 3*T at the end of the letter (lines 1065-1085).

This routine could form the basis of a program to generate automatic Morse code from messages typed in at the keyboard. However, writing the complementary program to receive and decode Morse is a much tougher problem, and this would provide an interesting challenge to any radio amateurs with micros.



N THIS, the final part of the camera, we first catch up on some Logic Board constructional notes. The f.e.t. TR20 (VN66AF) should have its heatsink doubled over, as shown in Fig. 3.1, to clear its headroom.

Scrape a small amount of track off the top-right hole where the coil is situated on the p.c.b. Link the coil to the casing as shown in Fig. 3.2. There are two links, and C50 (15µ/16V) to be mounted. The coil board is then mounted with C50 going to the lower part of the board (near IC3).

> Fig. 3.2. Coil sub-assembly

TR20

Fig. 3.1. Reshaping the heatsink of

FINAL ASSEMBLY INSTRUCTION

Having completed the construction of the individual boards of the camera, the final assembly consists of fitting the boards together and making the necessary interconnections. The vidicon tube and scan coil assembly is supplied complete with connecting wires to make assembly simpler.

The individual component boards may be sent for testing and setting up to Security Electronics and Engineering (Printoid Ltd.), 20/21 Alfric Square, Woodston, Peterborough. @ 07.33 329111. All components are available separately. Technical and supply problems should be sent direct to: PE/Seescan Camera Project, Security Electronics and Engineering, 20/21 Alfric Square, etc.



EP787

Connect the four boards together using the varelco connectors—these are polarised so that the connectors will only mate correctly in one position. Check that all connectors are making good contact.

Mount the inner rear panel to the rear of the camera chassis using four M3 6mm screws. Mount the co-axial socket on the inner face of the rear panel using two countersunk M3 screws, one tag washer and two M3 nuts. The tag washer fits on the screw in the red hole.

Using heavy gauge tinned copper wire, connect the coaxial socket core and the tag washer to the points indicated on the diagram of Fig. 3.4.

Connect the three wire plug from the video board to the logic board so that the black wire is near pin 8 or IC9 on the logic board.

Mount the tube/coil assembly onto the front of the camera chassis using four M3 20mm screws with M3 nuts, ensuring that the coil and tube are orientated as shown in Fig. 3.5.

Plug the scan coil connector onto the logic board so that the blue wire is nearest to the scan coils.

Fit the tube base connector to the tube ensuring that the short tube pin fits in the unnumbered hole of the connector.

Solder the earthing connections from the scan coil assembly to the earth rail on the video board, as shown in Fig. 3.6.

Solder the screened target connection to the video board as shown.

Screw the lens into the front of the camera. The camera chassis is now complete.

SETTING UP

The general description of the camera electronics, and of the methods of setting up have been left until last, as it was felt that it would be easier to visualise the camera as a whole once the individual sub-assemblies had been constructed. In making final adjustments to the camera, it is essential that a clear understanding of the function of each control has been achieved.

We hope to show how these units inter-connect and work together to form a high quality black and white television camera.

VIDICON TUBE

The inverter and high voltage power supply boards provide all the necessary dc voltages for the correct operation of the vidicon tube. When correctly adjusted, the vidicon will produce a voltage at its target ring which will be proportional to the intensity of light falling on the centre of the target window. In the absence of scanning wave-forms, the electron beam will produce a focussed spot near the centre of the target window. On no account should the tube ever be allowed to operate in this fashion as it will result in a spot being burnt onto the vidicon target.

If the video amplifier were connected to the target ring, it would amplify this smaller voltage, and an amplified signal would be fed to the UHF modulator. The purpose of the logic/scan board is twofold. First, it scans the electron beam across the target of the vidicon tube. Line scan moves the beam from left to right, field scan from top to bottom. These two scan waveforms are precisely locked together, and are carried out at the same speed as the electron beam moves across the monitor screen. Therefore, the voltage at the target ring of the vidicon will be continually changing—at any instant being proportional to the intensity of light at a particular spot over the surface of the vidicon target upon



Fig. 3.4. Connection to the VHF modulated video output socket



which the optical image of the scene is focussed. We now therefore have picture information available at the output of the video amplifier ready to send to the monitor. However, at the moment, the scanning of the vidicon tube is not linked to the scanning of the monitor cathode ray tube. We need to ensure that when the camera starts a scan at the top left hand corner of the picture the monitor is also scanning at the top left hand corner. To do this we send out a pulse (called a sync. pulse) at the start of each line and field. The monitor is forced to keep in step by these pulses, ensuring that a stable fully locked picture is obtained. Our line and field sync. pulses are obtained from the logic board, and are mixed with the video signal in the video amplifier. The waveform handled by the last two stages of the video amplifier, and then fed to the modulator is a composite containing both the video signal and synchronisation pulses. As mentioned in part four, blanking signals are also added to prevent the electron beam retracing its steps across the picture when it rapidly flies back to start the next line or field. These signals fed to the cathode of the vidicon tube cut off the electron beam between lines and at the end of fields preventing it reaching the rear of the target surface.

In testing the camera, it is important to ensure systematically that each of the sub-assemblies are correctly working.

PRESET CONTROLS	
Logic Board	Master clock oscillator
	Magnetic focus
Video Board	Automatic light control
and the second	Pedestal Sync
	DC level
High Voltage Power Supply Board	Target
	Beam
	Electrostatic focus

SETTING UP

Before applying power, turn all presets to midway position, with the exceptions of beam current, target voltage and automatic light control. The beam control should be adjusted for maximum negative voltage on its wiper, the target control turned fully anticlockwise until zero volts is on the wiper. The automatic light control should be fully anticlockwise. With these settings, the vidicon tube is fully cut off, and it will not be possible to damage the face of the tube in the event that the scan waveforms are incorrect.

Switch camera on; line frequency whistle should be heard immediately from the inverter transformer; if not, switch off and recheck connections. Set the height and width controls to maximum, i.e. fully clockwise. At this stage, if the camera is connected to a TV set tuned to channel 36 a pattern of lines should be seen on the screen. Adjust the master oscillator control to stabilise the pattern as far as possible. If no pattern is visible, just a blank, noise free raster, try altering the position of the pedestal black level control. Place a finger on the target connection; a great deal of noise (random patterns) should be visible on the TV screen. Adjust the pedestal control for the strongest display. Use an oscilloscope to check that correct scan wave forms are present at coil plug. See Fig. 2.11.

Set beam current control to give 30V at its wiper, and target control to $\frac{3}{4}$ maximum. By adjusting the beam control carefully, it should now be possible to detect some sensitivity to light when the camera lens is wide open. Now set the camera lens to about f.8 in normal illumination, then adjust the electrostatic focus control to give the crispest image. Aiming at an object 2 metres away, with a lens set at 2 metres, slacken off the tube retaining clamp and slide the tube carefully backwards and forwards to obtain the sharpest focus. Adjust the height and width controls until the edges of the picture just exclude the target windows; readjust the electrostatic focus, and beam controls, for the best picture. Trim the pedestal control for best picture with no crushing of dark greys into blacks. Adjust sync. control for best contrast and stability; or use oscilloscope and set up video output wave-form to CCIR standard. Adjust all the presets one at a time for optimum picture quality. Finally, advance the automatic light control preset until it just has an effect on the picture; back off this setting slightly. This is the correct setting for normal use.



JAPANESE IDEAS

The Nissan Motor Company of Japan has been toying with the idea of building a factory in Britain to make cars. Already the company is filing patents in Britain. The latest (British patent application 2 074 313) suggests an interesting new application from Nissan for light links.

Currently any switches on the steering wheel of a motor car, for instance the horn. button, produce electrical control signals. These are fed to the main car wiring harness via electro-mechanical slip rings between the moveable steering wheel and stationary steering wheel column. But slip rings suffer from wear. Also the contacts are bulky, and this limits the number of connection paths and thus the number of

emerging from the ring are picked up by photo-diode 24 which is secured to the steering wheel column. The electrical output diode 24 is amplified at 25, shaped at 26, decoded at 27 and fed to drive circuit

enables a full message to be transmitted in 15 millisec. Because the switch is likely to be depressed for at least 100 millisec, each coded message is transmitted at least twice for safety.



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Semiconductor UPDATE FEATURING LM1035 99000 SERIES 8X60 R.W. Coles

REMOTE CONTROL

In days of vore, when I ran a sort of audio amplifier assembly facility in my back shed, I became well versed in the tricky business of connecting sensitive volume, bass, and treble pots into the rest of the circuitry using carefully screened and earthed leads. Of course, these amplifiers used transistors or even valves (Ah! Pentodes!) which always needed an element of black-magic to ensure success anyway, but in today's chip orientated circuitry, electronic sorcery at the soldering iron stage seems out of place somehow. My own special witches' brew for amplifier control connections has been rendered obsolete overnight by a new chip from National, the LM1035.

The trouble with the old method of connecting amplifier controls was the fact that the pots had to be on the front panel while the amplifier itself was several inches away, and the interconnections made marvellous transmitting or receiving aerials for crosstalk, noise, mains hum, and even Radio Luxembourg. What National have done is to provide electronic control circuits which are controlled directly by d.c. voltage levels, so that the sensitive audio signal path never has to leave the chip. Pots are still used of course, but now they are used to get, not a signal level, but a d.c. level which can be slugged with a big electrolytic capacitor to remove any stray signals which could interfere. Inside the chip are two signal channels (for stereo) which have individual electronic attenuators for gain and balance, and individual voltage controlled filters for treble and bass control with each feature being controlled by a single voltage level so that there is no longer a need for expensive ganged pots!

The LM1035 has a distortion level of 0.02% at 1kHz and gives a volume control range of 80dB with 22dB of balance range. Maximum bass and treble boost or cut is 15dB, channel tracking is good to within 1dB and channel separation is -60dB at 1kHz. An on-chip voltage regulator is provided to allow the device to operate on unstabilised supply rails of from 8 to 20 volts, and the whole thing lives in a tiny 20 pin package.

SUPERCHARGED 9900

When Texas Instruments introduced the first true 16 bit microprocessor, the 9900, they claimed that it spelled "The end of the two-bit eight-bit" and sat back to watch their tall-in-the-saddle processor steal sockets from Intel and Motorola who were then still clinging to their puny 8 bit devices. Unfortunately the market was not really ready for a 16 bit chip, and although the 9900 was reasonably successful, it certainly did not sweep the other 8 bit chips off the board, and even found Itself a poor second when Intel, Zilog, and then Motorola introduced their own 1000 series 16 bitters several years later.

Well Texans don't take a licking easily, and so after a long cool analysis of the opposition, Texas have supercharged their 9900 design and brought out a new bit slinger family which appears to have the drop on those upstarts from the competition. Their new chip is still a sixteen bitter and is downwards compatible with the instruction set of the 9900, but there the similarity ends because the 99000 series offers some real innovations which increase its speed and power far beyond that of its ancestor and even beyond the fast guns of the Motorola 68000. First to hit the streets will be the 99105 which has 82 basic instructions including multiprecision arithmetic, stack operations, memory bit manipulations, and parallel I/O instructions to overcome the limitations of the weird serial-only scheme of the 9900. Next will be the 99110 which has the great additional feature of an on-chip floating point arithmetic library which allows it to add, subtract, multiply, divide, move and convert real variables in the IBM format, a feat which has hitherto required substantial external software.

Fundamental to all members of the new processor family will be a clock speed of 24 MHz (almost v.h.f. 1) which is twice as fast as the competition and means that with a memory cycle time of just 167 nS it will be necessary to choose the faster memory chips to take advantage of the power available. Inherent in the architecture of the family is a main memory address range of 256K bytes, and a separate "Macro-Store" address range of 120K bytes which can be used to augment the instruction set of the basic processor without the need to use main memory. The 99110 has its floating point arithmetic instructions implemented in Macro store space and actually uses a special on-chip 1K ROM and 32 byte RAM for this purpose. In the future other family members will use this same on-chip store to provide other high level features such as a real-time operating system or special I/O control routines. Unlike the 9900 which used a monster 64 pin package, the first members of the 99000 family both use 40 pin packages and have multiplexed data and address buses to suit.

FIFO

If you have a pet micro with a printer you probably get frustrated waiting for it to output listings at a rather pedestrian rate, during which it studiously ignores its keyboard and refuses to do anything else. One way round this problem is to create a spooler program which runs under interrupt control to give apparently simultaneous service to both the printer and the keyboard so that you can go on typing in more code while the listing is printed. This scheme relies on the fact that it takes a short time to send a character to the printer but it takes a much longer time for the printer to actually print it. Since the micro is only involved in the sending operation the spare time can be used to service other tasks.

Unfortunately, all the spooler programs I have seen implemented on micros have suffered from shortcomings such as missed keyboard characters while printing, and of course there is the complication of running under interrupt control which needs careful attention while writing programs. A much simpler solution is possible which is used already on many printers such as the Nascom Imp and involves the use of a character buffer store implemented in hardware and not under the control of the microcomputer itself. In this case the micro "sees" an apparently very fast printer and dumps whole pages of text in next to no time. If the listing will fit entirely within this print buffer then no waiting is involved, and with memory prices dropping all the time a big print buffer looks economically feasible.

The buffer can be fitted between a micro and a printer, and needs to be organised as a FIFO (First In First Out), which raises the problem of how to control the two memory address pointers required to control writing at one rate and reading at a different rate. The Nascom printer uses a microprocessor to control this feature, but this starts to look a bit complicated for a stand alone buffer unit which is why, to my knowledge, no designs have been published. Fortunately the concept has now become much more practical thanks to a new chip from Signetics called the 8X60 FIFO controller.

This 28 pin device contains all the arbitration and control logic and two 12 bit address pointer registers to handle a buffer RAM array of up to 4K characters, quite adequate for the majority of day to day printouts. If you used a couple of 2K by 8 static RAMs you could make the heart of a print buffer with just three chips, with additional devices probably required to handle interfacing to the micro and the printer. The exact form of any interface would depend upon whether serial or parallel drive is required. In the case of a serial in-serial out link two UARTs would be needed, but a Centronics parallel interface would be much simpler.

In addition to Read and Write strobe inputs and address outputs, the 8X60 also has outputs to say whether the buffer is full, half full, or empty. in Radio, Television & Electronics ICS have helped thousands of embitious people to move up into higher paid more secure jobs in the field of electronics - now it can be your turn. Whether you are a newcomer to the field or already working in the industry. ICS can provide you with the specialised training so essential

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Protection: Load line, momentary short circuit (typically 10 sec). Slew rate 15V/ μ s Rise time: Su6. S/N ratio 100db, Frequency resconse (-3dB):15H2-50kH2, Input sensitivity 500mV rms, Input impedance 100kD, Damping factor (82/100H2)>400. ILP Electronics Ltd, Freepost 2, Graham Bell House, Roper Close, Canterbury CT2 7EP, Kent.

HEAVY DUTY with heatsinks

Model No.	Output power Watts rms	DIST T.H.D. Typ at 1kHz	ORTION I.M.D. 50Hz/7kHz 4.1	Supply voltage Typ/Max	Size mm	Wi gms	Price Inc VAT	Price ex. VAT
HD 120	60w/4-8Ω	0.01%	<0.006%	±35±40	120 × 78 × 50	515	£25 85	£22:48
HD 200	120w/4·8Ω	0.01%	<0.006%	±45±50	120×78×60	620	£31 49	£27.38
HD 400	240w/4Ω	0.01%	<0.006%	±45±50	120×78×100	1025	£44 42	£38.63
								-

HEAVY DUTY without heatsinks

4



Protection: Load line, PERMANENT SHORT CIRCUIT (ideal for disco/group use should evidence of short circuit not be immediately apparent). The Heavy Duty range can claim additional output power devices and complementary protection circuitry with performance specs as for standard types.

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TIME FOR DATA PROCESSING

A major problem has been forcing its attention upon the groups who examine and analyse the data from satellites and probes. As usual the basic delay is funding. The amount of data has been so great that a very systematic priority clamp has had to be imposed upon some important sections of this work. It is of course readily understood that certain political requirements are involved and what might be considered by many of us as of humanitarian areas of priority are only of secondary consideration. With the present recession brought about indeed by the wrong priorities in many cases it could be deemed prudent to take an honest look at the terms of priorities. This requires a state of mental honesty extremely difficult to achieve when judged by action. A great many researchers all over the globe are seeking ways to improve the lot of mankind but work under conditions that demand for the continuance of their activities a constant campaign of begging donations from private resources for the funds to continue. So the inverse spiral of illogical thought publicly eulogises the endeavours of those engaged in these humanitarian works but denies them at the same time the means to forward them.

Now is the time to remember that the whole Earth and its environs is a system which is self generating or perhaps more correctly, regenerating and that every individual part is also the whole. Only now are some of the answers beginning to emerge. Much of the data could come from the environment of the atmosphere. Many false notions are still being bandied about such as that increasing carbon dioxide in the atmosphere will sooner or later make life impossible on Earth. Strange ideas about man's destruction of his environment AND EVEN HIMSELF is a tribute to vertical thinking and dealing in doom-mongering is an idiot course of action. Lateral thinking and a proper assessment with complete mental honesty in every phase of human activity is required to bring some impetus to those with vision and a certainty that the resources of the Earth are for the people of the Earth and should be brought into active use for their wellbeing. The mass of data requires funding and its ultimate benefit can be decided by the stroke of a pen in a political hand, which could deny the benefit of health to millions.

HIGH ALTITUDE EXPLORER SATELLITE

Explorer DE-1 has brought details of the aurora and the flow of solar energy and other matter from space. The interesting thing about the results provided by DE-1 and its companion DE-2 in a lower orbit which were equipped with instruments to provide data about the electric currents, fields and plasma between the magnetosphere, the ionosphere and the atmosphere is that it repeats history in the sense of 'it's been done before'. It is perhaps pertinent to point out that it is some fifty years since the basis of these phenomena were laid down and the archives contain records of papers on the subject of aurora.

In the latter part of the 1920's before the popularisation of the superheterodyne receiver multistage high frequency 'front ends' were used. Sometimes there were up to four stages. The coupling between stages was by HF transformers which looked like balls of string. Some of us searched in vain for the solution to what appeared to be instability. It appeared spasmodically and took the form of whistles sometimes rising in cadence and in other cases falling. Eventually there appeared headlines in the press to the effect that electrons shuttled backwards and forwards between the auroral points because they were trapped in the Earth's magnetic field. It was announced with some gusto that by timing some of these they followed lines of magnetic force more than 800m from the earth. These whistlers became the happy hunting ground for many postgraduates. Some of these 'whistlers' were given names like 'nose, lips, profile etc.' based on the shape which the graphs took as a result of the changing frequency. It would be interesting to know how many readers remember those days.

Now all this has been brought up to date as a result of DE-1 and DE-2 activity. The orbits of the satellites are:

DE-1 an eccentric orbit apogee 15,000 mi. perigee 354mi.

DE-2 apogee 628mi. perigee 192mi.

The arrangement of these two spacecraft allows the bracketing of the region of the leaking of low energy plasma into the magnetic field cusp from space where it is transformed into the high energy stream which initiates the aurora. It is well known now that the aurora is formed at a level which varies from 3,000 to 12,000 miles several times in 24 hours. At these levels the electrons trapped in the magnetic field are reflected back along the field lines and may continue this oscillating activity until their energy falls below a certain level.

The imaging system of DE-1 consists of three spin-scan photometers which permit daytime viewing of the aurora. The images are built up from line scans of the earth at the rate of one line per rotation of the spin stabilised satellite. This system uses 'super-reflecting' mirrors to avoid the light scatter which prevents the use of lenses for this work. Two cameras operate in the visual part of the spectrum and a third operates in the ultra-violet. This imaging system is being used for other tasks also which include ozone observations, a special search for marine bioluminescence from space and the observation from high polar altitudes of the Earth-glow which is now called the geo-corona.

As the DE-1 apogee moves towards the Earth's equatorial region it is hoped that the cameras will be able to observe the red-glow bands that circle the Earth to the north and south of the equator. The drifting of the apogee will later enable the south polar regions to be observed. DE-2 carries instrumentation for the observation of particle and wave field environments. The instruments are capable of providing details of electron velocity every 2km.

A report by S. D. Shawhan on the radio waves generated in the auroral regions revealed that the more powerful emissions are associated with auroral storms. The frequencies are of the order 100-200kHz though sometimes as high as 800kHz (this was the range of whistlers). They are directed away from Earth and have powers of the order of billions of watts and can be picked up more than 15,000 miles from the auroral zones. This is of the same order as the radiations from Jupiter. They are characterised by rising and falling tones, each burst lasting about a second and changing in frequency by 1kHz. Detailed measurements of these waves have been made by DE-1. Its plasma instrument which monitors frequencies from a few Hertz to 400MHz detects the most intense waves when the spacecraft crosses magnetic field lines that pass through the aurora. Waves were detected by both satellites in the audio range, that is below 20kHz. This suggests that these waves must be guided away from the power source by the Earth's magnetic field.

High winds and high temperatures have also been recorded by DE-2. Data reveals that there are winds in the auroral zone which are of the order of hundreds of miles per hour and at times over 1,000 miles an hour. Temperatures have reached the order of 1,500 to 2,000K.

It is not possible to resist commenting that the Earth is far from approachable by alien spacecraft.

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The Boeing Aerospace Company has been selected to conduct a one year study for the National Aeronautics and Space Administration of an unmanned launch vehicle.

The vehicle which is designated SRB-X is being considered as a means of augmenting the space shuttle when payload and space is not immediately available in the shuttle orbiter. It will be capable of boosting a 65,000lb payload into low earth orbit or 12,000lb into a geosynchronous orbit. The SRB-X would be built from tried components with minimum modifications to boosters and launch facilities.

There are also three other studies—a liquid rocket booster and a side mounted shuttle derived cargo vehicle, studied by Martin Marietta; and another shuttle derived cargo vehicle—in line version, under study by Boeing.



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M05 200	120w/4-812	<0 005% <0 006%	+55+60	120 × 78 × 80	850	£38 48	£33.46
M0S 400	240w/49	<0.005% <0.006%	±55±60	120 × 78 × 100	1025	£52 20	£45 39

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MOS 200P	120w/4-812	< 0 005% < 0 006%	±55±60	120×26×80	420	£32 81	£28 53
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INDEX TO ADVERTISERS

Ace Mailtronix Adam Hall Supplies	74 77	Litesold London Electronics College		73 77
Ambit	14	Maplin Electronics		Cover IV
Autumn Products	72	Microstate Ltd Midwich Computer		76
Barrie Beckman Instruments	78 27	Millhill Modern Book Co		74 79
Bi-Pak Blackstar	8,9	Musicraft		80
School School	4	Parndon P.C. Electronics		74 76 71
Cambridge Kits Cambridge Learning	78 10	Pimac		6
Chordgate Clef Products	72	Plessey Radar Powell T.		72 80
C.R. Supply Co Crofton Electronics	76 14,69	Powertran Printoid Ltd	Co	ver II, 14 6
C.U.A	10	Printronix Proto Design	·	78 78
Dataman Design	74	Radio Components Specialist	s	79
Electronize Design	55 15 74	Riscomp Ltd.	••••	16
Gemini	75	Scientific Wire Co Sinclair Research		78 12, <u>13</u>
Global Specialities Corporation	11	Solid State		75
Heathkit	39	Tempus	80,	Cover III 70
House of Instruments	43	Videotopo		5
ICS Intertext ILP Electronics 65, 67, 69, 71, 7	67 73, 7 5	The Vintage Wireless Co.		76
JSO Studios	77	Watford Electronics William Stuart Systems (Big I William Stuart Systems	Ears)	2,3 77
Knight A	76	Scope Wilmslow Audio		78 10

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