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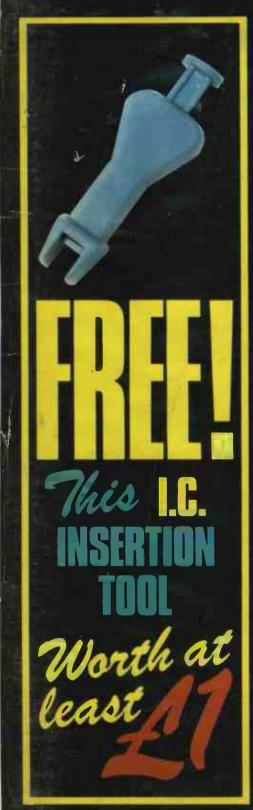
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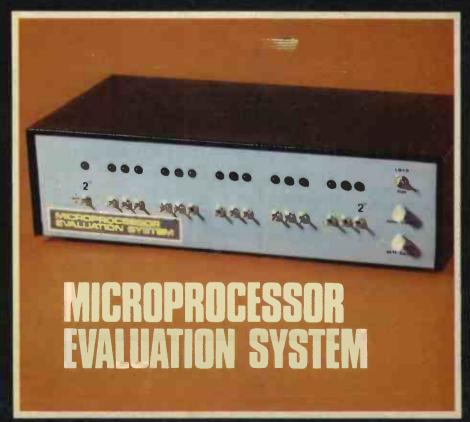
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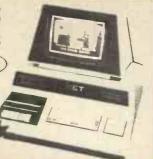
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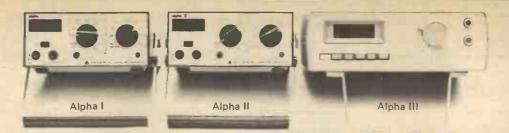
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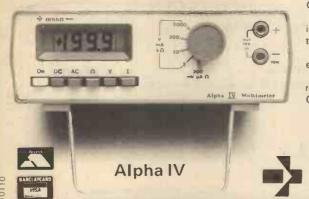
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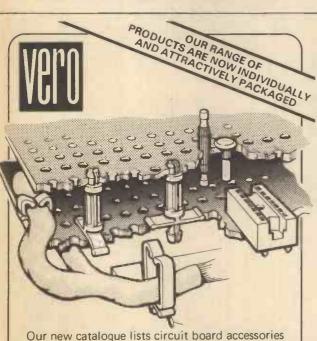
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LS26 48 LS114 50 LS183 298
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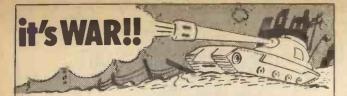
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7493	32	74193	98	4054	128	4512	98	LM
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Practical Electronics May 1979



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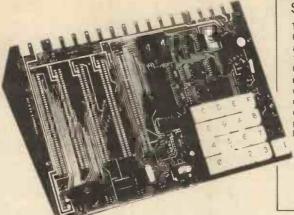
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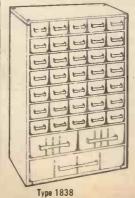
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QUARTZ LCD Alarm Chronograph with Dual Time Zone Facility

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Perpetual calendar, day, date, month and year. 24 hour alarm with on/ off indication, 1/10 second chronograph measuring net, lap and first and second place times. Dual time zone facility night light. Only 9mm thick.

£27.65



QUARTZ LCD Alarm Chrono with front alarm

Dual time Ten function, 6 digit. Hours, mins., secs., date, day of week, stopwatch, split time, alarm, second watch (dual time), backlight.

£29.65



M7

QUARTZ LCD Alarm Chrono

Ten function, 6 digit. Hours, mins., secs., date, day of week, stopwatch, split time, alarm, second watch (dual time), backlight. FRONT BUTTON

OPERATION.

£29.65



M8

SOLAR QUARTZ LCD Chronograph

6 digit, 11 function. Hours, mins., secs. 1/100, 1/10 secs., mins

Split and lap modules. Auto calendar and back light. Powered from solar panel with battery back-up.

£15.95



SEIKO Alarm Chrono

LCD, hours, mins., secs., day of week, month, day and date, 24 hour Alarm, 12 hour chronograph, 1/10th secs., and lap time, Back light, stainless steel, HARDLEX glass. List Price £130.00 METAC PRICE

£105.00



M10

SEIKO Chronograph

LCD, hours, mins., secs., day of week, month, day, date, 12 hour chronograph, 1/10th secs and lap-time. Back light, stainless steel water resistant. HARDLEX glass.

List Price £85.00 METAC PRICE

£68.00

M11

10:08 42

SOLAR QUARTZ LCD 5 Function

Genuine Solar

Solar panel with battery back-up. Back light and auto calendar. Hours, mins., secs. day, date. Quality metal bracelet.

£ 9.95

Guaranteed same day despatch

M12

130

HANIMEX **Electronic** LED Alarm Clock



reauries and Specification:

Hour/minute display. Large LED display with
p.m. and alarm on indicator, 24 Hours alarm with
on/off control. Display flashing for power loss
indication, Repeatable 9-minute snooze. Display
bright/dim modes control. Size: 5.15" x 3,93" x
2,36" (131mm x 11mm x 60mm).

Weight: 1.43 lbs (0.65 kg). Features and Specification:

£8.65

Guaranteed same day despatch.

QUARTZ LCD Ladies Slim Bracelet

Hours, mins., secs., day, date and back light and auto calendar. Elegant metal bracelet in silver or gold.

State preference £15.95

Guaranteed same day despatch.



M14

QUARTZ LCD Ladies 5 Function

Only 25 x 20mm and 6mm thick. 5 function. Hours, mins., secs., day, date and back light and auto calendar. Elegant metal bracelet in silver or gold. State preference.

£9.95

Guaranteed same day despatch



M15

DIGITAL LED CLOCK



Automatic brightness control. Weekend alarm cancel.

Features and Specification

reatures and spectrication:

Hour/minute display, Large LED display with p.m. and alarm on indicator, 24 Hours alarm with on-oft control. Display flashing for power loss indicator. Repeatable 9-minute snooze. Automatic brightness control. Weekend alarm cancel.

£10.95

M16

HOW TO ORDER

Payment can be made by sending cheque, postal order, Barclay, Access or American Express card numbers. Write your name, address and the order details clearly, enclose 30p for post and packing or the amount stated. We do not wait to clear your cheque before sending the goods so this will not delay delivery. All products carry 1 year guaranteee and full money back 10 day reassurance, Battery fitting service is available at our shops. All prices include VAT.

Trade enquiries: Send for a complete list of trade prices - minimum order value £100. Telephone Orders: Credit card customers can telephone orders direct to Daventry or Edgware Rd., 24 hour phone service at both shops: 01-723 4753 03272-76545.





CALLERS WELCOME Shops open 9.30 - 6.00.

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M13

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Overall size 12In wide x 8In deep x 2½In high.
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ı			AVAILABLE	. 013	SENIES	_	IN FARAL	LEL CON	INECTION		
ı	Туре	Voltage	Curent	£	p/p	۱	Туре	Voltage	Current	£	p/p
	06FE06 08FE06 12FE06 20FE06	6+6 6+6 6+6	0.5A EACH 0.6A EACH 1A EACH 1.6A EACH	1.90 2.10 2.74	50p 50p 60p 70p		20FE24 50FE24 60FE24 80FE24	24+24 24+24	0.4A EACH 0.8A EACH 1.2A EACH 1.5A EACH	3.25	70p 70p 85p 100p
	50FE06 60FE06	0.0	3A EACH	3.98	70p 85p		50FE28 60FE28 80FE28		0,75A EACH 1.1A EACH 1.4A EACH	3.98	70p 85p 100p
	06FE09 08FE09 12FE09 20FE09 50FE09 60FE09	9+9 9+9 9+9 9+9 9+9	0.3A EACH 0.5A EACH 0.75A EACH 1A EACH 2.5A EACH 3A EACH	1.90 2.10 2.74 3.25	50p 50p 60p 70p 70p 85p		20FE30 50FE30 60FE30 80FE30	30+30 30+30 30+30	0.35A EACH 0.75A EACH 1A EACH	2.74 3.25 3.98	70p 70p 85p
ı						l			RANGE, VO		
	06FE12 08FE12	12+12 12+12	0.25A EACH 0.3A EACH	1.90	50p 50p			12-0-1	5, 6, 8, 9, 10. 12 OR 15.0-15	12, 1	5, 18,
	12FE12 20FE12	12+12 12+12	0.5A EACH	2.74	60p 70p		30FE30	0-12-15 24-30	1A	3.55	70p
ı	50FE12 60FE12 80FE12	12+12 12+12 12+12	2A EACH 2.5A EACH 3A EACH	3.98	70p 85p 100p		60FE36	0-12-15 24-30 0-12-15	2A	4.78	85p
ı							80FE36	24-30	3A	5.95	100p
	06FE15 08FE15	15+15 15+15	0.2A EACH 0.25A EACH	1.90	50p 50p		100FE40	0-12-1 5 24- 3 0		7.10	115p
ı	12FE15 20FE15	15+15 15+15	0.4A EACH		60p 70p	Ŀ	С	ENTRET	APSECOND	ARY	
ı	50FE15	15+15	1.6A EACH	3.25	70p	ı	FE06	6-0-6	1A EACH	12.10	60p
ı	60FE15	15+15 15+15	2A EACH 3A EACH		85p 100p	۱	FE09	9-0-9	1A EACH	2.74	70p 70p
ı					ТООР	ı	FE15	15-0-15	1A EACH		70p
I	06FE20	20+20	0.15A EACH		50p	I		20-0-20		3.25	70p
ı	08FE20 12FE20	20+20	0.2A EACH		50p 60p	1		26-0-26 28-0-28			100p
ı	20FE20	20+20	0.5A EACH	2.74	70p	ı	60FE60	30-0-30	1A EACH	3.98	100p
ı	50FE20 60FE20	20+20	1.2A EACH 1.5A EACH		70p 85p	1	100FE26			5.15	115p
ı	80FE20	20+20	2A EACH				100FE36				
	CH	ARGER	TRANSFOR	MER			ALP CO	ED ALID	O CROSS O		
ı	48FE12			3.25	70p		FEO1		IO CROSS O	0.26	20p
ı	66FE12	0-6-12	5A 6A	4.00	85p 100p		FE03	0.3mH		0.26	20p
п	/UFE IZ	0-0-12	U.A.	0.10	. OOP	rii I	FFOR	0.5mH		0.30	20n

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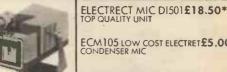


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Available complete and ready to plug in or as an easy to connect module with all controls except monitor switch already fitted—full instructions supplied.

FEATURES INCLUDE: Twin Deck – Mic & Tape Inputs – Wide range bass & treble controls – Full headphone monitoring – Crossfade – Professional stan-dard performance.

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PDF100 Reflex Bin - Twin Horns - Integrated Slave Amplifier - Accepts mono or stereo signals

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 ☐ Send for details
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Quality audio modules and accessories for

S450

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Fitted with
phase lock-loop





FREQUENCY RANGE	88-108 Mhz
SENSITIVITY	3 0 µV
BANDWIDTH	250 kHz
SPURIOUS REJECTION	50 dB
SELECTIVITY ± 400 kHz	55 dB
AUDIO OUTPUT (22:5 kHz deviation	1) 100 mV
STEREO SEPARATION	30 dB
SUPPLY REQUIREMENTS	20 to 30V (90mA max)
AERIAL IMPEDANCE	75 ohms
DIMENSIONS	240mm - 110mm - 32mm

The 450 Tuner provides instant programme sefection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be altered as often as you choose, simply by changing the settings of the pre-sel controls. Features include FET input stage. Vari-Cap diode tuning. Switched AFC LED Stere Indicator.

Stereo 30 COMPLETE AUDIO £19-18 + 66p p&p + 121" VAT 7 + 7w R.M.S

OUTPUT POWER	7 Watts RMS
LOAD IMPEDANCE	8 ohms
TOTAL HARMONIC DISTORTION	Less than 5% (Typically 3%)
FREQUENCY RESPONSE	50 Hz to 20 kHz ± 3dBs
TONE CONTROL RANGE	± 12 dBs at 100Hz and 10kHz
SENSITIVITY	190 mV for full output
INPUT IMPEDANCE	1 M ohms
TRANSFORMER REQUIREMENTS	22 V.A.C. rated at 1A
(Less controls and panel)	200mm - t30mm - 33mm

The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supply. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality ceramic pick-up, stereo tuner, stereo tape deck etc. Simple to install, capable of producing really first class results, this unit is supplied with full instructions, black front panel, knobs, main switch, luse and fuse holder and universal

25w R.M.S

AL60

AUDIO AMPLIFIER

MODULE 25 Watts RMS £4-69 + 35p p&p 121% VAT

	OUTPUT POWER	25 Watts RMS
-	SUPPLY	30-50 V
	LOAD IMPEDANCE	8-16 ohms
-	TOTAL HARMONIC DISTORTION	Less than 1% (Typically 06%)
	FREQUENCY RESPONSE	20 Hz to 30 kHz + 2 dBs
- 1	SENSITIVITY	280 mV for full output
1	MAX. HEAT SINK TEMPERATURE	90°C
П	DIMENSIONS	103mm - 64mm - 15mm

This high quality audio amplifier module is for use in audio equipment and slereo amplifiers and provides output powers up to 25 RMS with distortion levels below 0.1%.

35w AL80 AUDIO 4 AMPLIFIER MODULE £7.34*

	OUTPUT POWER	35 Watts RMS
١	SUPPLY	40-60 V
ı	LOAD IMPEDANCE	8-16 ohms
1	TOTAL HARMONIC DISTORTION	Less than 1% (Typically 06%)
ı	FREQUENCY RESPONSE	20 Hz to 30 kHz × 2 dBs
ı	SENSITIVITY	280 mV for full output
ł	MAX. HEAT SINK TEMPERATURE	90°C
ı	DIMENSIONS	103mm · 64mm · 15mm
-		

The AL80 is similar in design to the AL80 above and is of the same high quality but provides output powers up to 35W with distortion levels below 0.1%

125W R.M.S. **AL250**

OUTPUT POWER	125 Watts RMS continuous
OPERATING VOLTAGE	50-80 V
LOAOS	4-16 ohms
FREQUENCY RESPONSE	25 Hz 20 kHz measured at 100 Watts
SENSITIVITY FOR 100 WATTS O/P AT 1 kHz	450 mV
INPUT IMPEDANCE	33 K ohms
TOTAL MARMONIC DISTORTION 50 WATTS into 4 ohms 50 WATTS into 8 ohms	0·1% 0 06%

f17.82* +66pp&p + 1% VAT

This unit, designated AL250, is a power amplifier providing an output of up to 125W RMS, into a 4 ohm load

AL30A AUDIO AMPLIFIER MODULES



Ì	MAXIMUM SUPPLY VOLTAGE	30 V
9	POWER OUTPUT for 2% THD	10 Watts RMS
ı	TOTAL HARMONIC DISTORTION	Less than 25%
3	LOAD IMPEDANCE	8-16 ohms
ı	INPUT IMPEDANCE	100 K ohms
1	FREQUENCY RESPONSE	50 Hz-25 kHz ± 3 dBs
1	SENSITIVITY	75 mV for full output
ı	DIMENSIONS	74mm - 63mm - 28mm

These low cost 10 watt modules offer the utmost In reliability and performance, whilst being compact In size.

SPM80

£16-05

+ 121% VAT

STABILISED POWER SUPPLY £4.40 + 350 pap + 121% VAT



и		
	INPUT A.C. VOLTAGE	33-40V
	OUTPUT D.C. VOLTAGE	33 V nominal
	OUTPUT CURRENT	10 mA-1:5 amps
	OVERLOAD CURRENT	1.7 amps approx.
	DIMENSIONS	105mm + 63mm > 30mm

Designed to power two AL60s at 15 Walts per channel simultaneously. Circuit Techniques include full short circuit protection.



ì	FREQUENCY RESPONSE	20 Hz to 20 kHz × 1 dB
	TOTAL HARMONIC DISTORTION	Less than 1% (Typically 07%)
	SENSITIVITY 1. TAPE INPUTS 2. RADIO TUNER 3. MAGNETIC P.U.	100 mV/100 K ohms } For an 100 mV/100 K ohms } output 3-5 mV/50 K ohms] 250 mV
	EQUALISATION	Within ± 1 dB from 20 Hz to 20 kHz
	BASS CONTROL RANGE	± 15 dBs at 75 Hz
	TREBLE CONTROL RANGE	+ 10-20 dBs at 15 kHz
	SIGNAL/NOISE RATIO	Better than 65 dBs (All Inputs)
	INPUT OVERLOAD	Better than 26 dBs (All inputs)
	SUPPLY	20 to 40 V
	DIMENSIONS	300 × 90 × 33mm (less controls)

A top quality stereo pre-amplifier and tone control unit, the PA100 provides a comprehensive solution to the front end requirements of stereo amplifiers or audio units. The six push button selector switch gives a choice of inputs together with two filters for high and low frequencies.

MAGNETIC CARTRIDGE PRE-AMPLIFIER



Enjoy the quality of a magnetic cartridge with your existing ceramic equipment using the MPA 30 which is a high quality pre-amplifier enabling magnetic cartridges to be used where facilities exist for the use of ceramic cartridges only. +35p p&p + 12;% VAT

SENSITIVITY	3 5 mV for 100 mV output
EQUALISATION	. Within ± 1 dB from 20 Mz to 20 kHz
INPUT IMPEDANCE	50 K ohms
SUPPLY	18 to 30 V—re earth
DIMENSIONS	110 · 50 · 25mm (inc DIN socket)

PA12 STEREO PRE-AMPLIFIER



£7.78

The PA12 Stereo Pre Amplifier chassis is designed and recommended for use with the AL 20/30 Audio Amplifier Modules, the PS12 power supply and the T538 Transformer. Features include on/off volume, Balance, Bass and Treble controls. Complete with tape output.

tupe odiput.
20 Hz-20 kHz (-3dB)
± 12 dB at 60 Hz
± 14 dB at 10 kHz
1 Meg. ohm
300 mV
60 dB
65 dB
± 20 dB
25 K ohms
152mm / 84mm · 25mm

PS12 POWER SUPPLY MODULE

Power supply for AL20A-30A, PA12, S450 etc. Transformer T538

Input A.C. Voltage 15-20V. Output D.C. Voltage 22-30V approx. (Dependent upon input.) Output Current 800mA

maximum. Dimensions 60 x 43 x 26mm

£1.90 + 12 1% VAT + 35p p&p.

BP124 SIREN ALARM

American Police screamer powered from any 12 volt supply into 4 or 8 ohm speaker ideal for car burglar alarm, freezer break-down, and other security purposes

ONLY £3.50 + 8% VAT + 35p p&p.



5 WATTS -

TOBETH

MA60 HI-FI AMPLIFIER KIT

Build you own top quality amplifier, save yourself pounds. The MA60 hit comprises the following Bl-hits modules, 2 × AL60 amps, 1 × PA100 pre-amp, 1 × SPM80 stab. power supply, 1 × BM180 transl, giving 15 watts RMS per channel STEREO. All modules covered by the Bl-PAK salisfaction or money back guarantee. Details of the above modules are in this ad. Price £32 00 + 121% VAT + \$2p p&p.

TC60 KIT

A beautifully designed genuine TEAK WOOD veneered cabinet to put the professional touches to your home built amplifler. Full set of parts Incl. Front & Back Panels, Knobs, Chassis, Fuses, Sockets, Noen, etc. Ideal for the MA60. Size. 425mm - 290mm >

Price £19-95 + 12}% VAT + 86p p&p

TRANSFORMERS

T538 For use with S.450 AL30A MPA30
Order No. 2036
Price: £3·20 + 55p pap + 12½ % VAT
7205 For use with Stereo 30
Order No. 2050
BMT80 For use with AL60 SPM80
Order No. 2034
Price: £5·40 + 86p pap + 12½ % VAT
BMT250 For use with AL50
Order No. 2034
Price: £6·35 + £1·10 pap + 12½ °, VAT
Order No. 2035 2040. For use with AL60 Order No. 2040 Price £5-20 + 80p p&p + 121% V.A.T.

Order No. 2040 2041, For use with ALBO, AL120 and AL250 Order No. 2041. Price £6 80 + 86p p&p + 12 ½% V.A.T

TEAK 30, 32 x 23 x 8cm, designed mainly for use with our stereo 30 Audio System but has proved very helpful to home constructors. Fitted with solid uncut front and back, o/n 139, £5-95, + 12\frac{1}{2}\times V.A.T. p&p 70p.

TEAK 60, 42 × 29 × 9cm, for use with AL60/MK60 Audio Kit. Useful for the home constructor requiring an amplifier sleeve – has no front or back panel. o/n 140, £7.00. + 12 ½ V.A.T. p&p 85p.

Professionals and Enthusiasts from BI-PAK

AL120 AUDIO AMPLIFIER (With integral heat sink and

short-circuit protection £11.95



50 Watts R.M.S.
70 Watts
8-16 ohms
05% Max. (Typically 02%)
25Hz-20kHz
500mV
45 deg. C
192 x 89 x 49 mm

Introduced to fulfill the demand for a fully protected power amp., capable of driving high quality speaker systems at up to 50w with distortion levels below 05%. Ideal for domestic use, Discos, P.A. systems, electronic organs etc. The generously rated components ensure continuous operation at high output levels.

SPM120 STABILISEO POWER SUPPLIES SPW120/45 SMP120/55 SMP120/65 £5.80 + 121% V.A.T. P. & P. 35p

AC INPUTS	
SPM120/45	40-48v
SPM120/55	50-USv
SPM120/65	60-65v
OUTPUT CURRENT	2.5A
RIPPLE	1A 100mV 2A 150mV

SPM120 is a fixed voltage stabiliser available with an output voltage of either 45v, 55v, or 65v. Designed primarily for use in audio applications, the stabiliser which provides output currents up to 2.5A., operates direct from a mains transformer requiring only the addition of 2 Electrolytic capacitors to complete the s/c protection.

GE100 Mk2. LLUML 10 CHANNEL MONOGRAPHIC £20.00

Control Range	±12dB
Dynamic Range	110dB
Maximum Output	+15dB
Frequency Response	30Hz-20KHz (±1dB)
Power Supply	15-0-15v.
Voltage Handling Input	3v R.M.S.
T.H.D.	005%

Only 155mm x 65mm x 50mm including the 10 x 10K 1in slider potentiometers and knobs which are mounted on a board positioned above the circuitry. In the frequency range of 31Hz to 20KHz you can cut and boast ±12dB with the 10 sliders, each of which has its frequency marked on the circuit board. The GE100 has numerous uses including mixers, P.A. systems and discos. It will also greatly improve the sound reproduction of your existing audio equipment. Power Supply for GE100, o/d SG30 f 3.80.

VPS30 REGULATED VARIABLE STABILISED POWER SUPPLY £7.60 +8% V.A.T.



AC Input Maximum	25v
Voltage Regulation	, 2-30v
Regulated Current	0-2A
Incorporating short circuit pr	otection

This NEW versatile Regulated Variable Stabilised Power Supply with short circuit protection and current limiting, is a must for all electronics enthusiasts. It incorporates adjustable voltage from 2v 30v, with a current limiting range of 0–2A. With this module there is no need to build a separate power supply for each of your projects, with the simple addition of a transformer (ord 2031, 0–1ma (o/d 1310 or 1305), plus a suitable shunt, a voltmeter (o/d 1311 or 1306), a 470ohm pot fo/d 1896), a 4K7 pot fo/d 1899), it can be used again and again as a self-contained bench, power supply, eliminating the use of batteries and thus saving

SENSITIVITY

EQUALISATION

FREQUENCY RESPONSE

TOTAL HARMONIC DISTORTION

PA200 STEREO PRE-AMPLIFIER



JU	933	Ž,	1.	

£16.55		
+12½% V.A.T. P. & P. 40p		

43.22.2	BASS CONTROL MANGE	± 15085 at 75HZ
A = -	TREBLE CONTROL RANGE	+ 10-20d8s at 15kHz
	SIGNAL/NOISE RATIO	Better than 65dBs (All inputs)
£16.55	INPUT OVERLOAD	Better than 2dBs (All inputs)
121% V.A.T.	SUPPLY	35 to 706v
2. & P. 40p	DIMENSIONS	300 x 90 x 33mm (less controls
he PA200 is basically our popular PA10	00. Modifications have been made to make it compa	itible with the higher output AL120 and

HEADPHONES

A top quality headphone with cushioned earpads and headband. Separate balance/volume controls. Stereo or Mono switch. Impedance: 8 ohms. Frequency: 30-18,000Hz. o/n 884. £8.70. + 12 ½% V.A.T. p&p 70p. A brilliant compromise between price and performance. Superb stereo reproduction for the newcomer to Hi-Fi. Impedance 8 ohms. Frequency; 30-15,000Hz. o/n 885. £4.40. + 12 ½% V.A.T. p&p.50p.

HI-FI ACCESSORIES

Parallel Tracking GROOV KLEEN
The very latest in automatic record cleaning. Designed to suit all modern single play decks. Simple to fit, it is extremely efficient. Complete with two types of base and three height extensions. o/n 8101, £3.68, +8% V.A.T. p&p.35p.

Cassette Tape Editing Kit Enables cassette tapes to be edited and joined easily, quickly and accurately. Kit comprises: Tape Splicer (* (3.2mm). 2. Precision Tape Cutters. Tape Piercer. 9 Self-adhesive Labels. Reel of Splicing Tape. 3 Winders and removers and instructions. all in a handy wallet. o/n 811. £2.40. + V.A.T. p&p 35p.

GROOV-STAT
The BIB Groov-Stat static reducer neutralises the static charge on records and other plastic surfaces. o/n 8103.£5.45. + 8% V.A.T. p&p 35p.

Cassette Head Cleaner Essential for cleaning of tape heads, capstans and rollers. Pack contains Tape Head Applicator and tape head polisher tools. Plus bottle of special formula cleaning fluid and full instructions. o/n 832, £0-64. + 12 }% V A.T. p&p 35p.

METERS

1. TAPE 2. RADIO TUNER 3. MAGNETIC P.U.

Miniature Balance & Tuning Meter
Miniature moving-coil meter for stereo balance
indicator, tuning indicator for FM or similar
application. Pointer at centre indicates zero or null
position. Robust construction. Sensitivity:
100-0-100MA, Dimensions: 23 x 22 x 26mm.
o/n 1318.£1.95 +8% V.A.T. p&p 35p.



Balance and Tuning Meter Clear view edgewise meter. Centre zero application. Sensitivity: 100 0-100UA. Dimensions: 45 x 22 x 34mm. o/n 1319. £2.00. +8% V.A.T. p&p 35p.

20Hz to 20kHz × 1dB

100mV/100 K ohms 3.5mV/50 K ohms

Within ± 1dB from 20Hz to 20kHz

Less than 1% (Typically -70%)

100mV/100 K ohms For an

Miniature Level Meter
Moving coil, for accurate level indication for tape
recorders, amplifiers etc. Neat design, rugged construction will withstand five times rated value.
Sensitivity: FSD: 200UA. OdB. 130UA. Dimensions: 23 × 22 × 26mm. o/n 1320, £2.80. +8%
V.A.T. p&p 35p.



Vu Meter Calibrated -20 to +3 and 0-100%, making it suitable for use as a recording level meter or as a power output indicator. Sensitivity: 130uA. Dimensions; 40 x 29mm. o/n 1321. £2.00. +8%



AC-DC enables a large range of battery powered radios, recorders, calculators to be run off the mains. (220-240v AC) Switchable for 6.7.5 or 9 volts. Current rating 2,500mA. Polarity reversing switch Universal plug incorporated. o/n 137 £3-60. + 12 1 1 2 1 3 V.A.T. p&p 35p. DC DC for use in all cars, boats etc., with pos or neg, earth for a regulated output of 6, 2.5 or 9 volts DC at 1A max, For radios, recorders etc. o/n 138, £2.80. + 12,1% V A.T. p&p 35p.

CROSSOVER NETWORKS

2-WAY channels for high and low frequencies to correct speakers – high to tweeters, low to woofers. Complete with instructions. Frequency: 3,000Hz. o/n 1904.£1.10. + 12 \{\frac{1}{2}\% \text{ V A T. p&p 35p.}}

2-WAY for 8 ohm speakers up to 30 watts. Frequency: .3KHz. o/n 1905. £1.65, + 12,3% V.A.T. p&p 35p.

3-WAY for 8 ohms speakers up to 30 watts. Frequency: 800Hz and 4.5KHz. o/n 1906. €2.95. + 12 ½% V.A.T. p&p 35p.

MICROPHONES

DYNAMIC CASSETTE

OYNAMIC CASSETTE for equipment requiring a high quality microphone. Sturdy, solid moulded body in black with neat chrome surround. Pick-up pattern is omnidirectional. On/Off switch, 1 metre of tough lead with floating 2 5 and 3.5mm plugs. Matching moulded strut. Impedance: 200 ohms. Sensitivity, 90dB. Frequency; 90-10,000Hz. Size: 20mm dia x 120mm.o/n 1326. £1-60 + 12;% V.A.T. p&p 35p.

DYNAMIC MICROPHONE

Superior quality portable cassette recorder mike with built-in remote control switch and lead fitted with 5-pin 240° DIN plug fremote switch and 3-pin DIN plug Imicrophonel. Provides a direct replacement for those supplied with recorders. With detachable stand. Omnidirectional. Impedance: 200 ohms. Freq. response: 100 to 10.000Hz. Sensitivity: 79dB'at 1,000Hz. o/n 1327. £2.65. + 123/%. V.A.T. p&p 35p.

RE - 317: DYNAMIC MICROPHONE

ne = 317: UTNAMIC MICHOPHONE Highly sensitive, high grade desk or hand mike suitable for use with many popular cassette decks. Incorporates On/Off switch and 1 metre lead with moulded standard jack plug. Complete with desk stand Omnidirectional. Impedance: 5,000 ohms. Freq response: 100 at 12,000Hz. Sensitivity: [-7dB at 1,000Hz). o/n 1336. £400. + 12 \frac{1}{2}\%

OMNIDIRECTIONAL CARDIOID

OMNIDHECTIONAL CARDIOID

Ownered by a 13-b battery located within the aluminium body. Satin silver finish with front disk protection to the diaphragm housing on/Off switch. Also with Busby type windshied. "U" bracket and stem and extremely supple cable. Consumption. 0.2mA from 13-battery providing approx. 3-10.000 hours continuous life. Impedance: 600-ohms. Individual Size. 23mm dia + 267 mm. o/n 1329.2-12.80. 123% V.A.T. p6p 35p.

UNIDIRECTIONAL CARDIOID
Dual imp 600 and 50,000 ohms. Response 50 to 14,000Hz.
Sensitivity \$4d8 at 50K/ohms. Size: 1\frac{1}{2} dia x 6\frac{1}{2}" long. Weight approx.. 190gm. o/n 1328.£10.95 . 12\frac{1}{2}% V.A.T. p&p 35p.

GOOSENECK CHROME FLEXIBLE HOLDERS Length 320mm. o/n 1333. £2.40. + 12\% V.A.T. p&p 35p. Length 515mm.o/n 1334. £3.40. + 12\% V.A.T p&p 35p.

FLOOR STAND Heavy chrome. Stow-away feet with rubber ends for maximum stability. Oraws to a height of 5' maximum. o/n 1335. + £9-50. 124% V.A.T. p&p 85p.

BOOM ARM for use with the above stand. Heavy chromed metal, it gives 30" reach from the stand. o/n 1337 £9-20. + 123% V.A.T. p&p 70p.

WINDSHIELD COVERS

o/n 1331 Medium per pair £1.20. + 12½% V.A.T. p&p 35p. o/n 1332 Large per pair £1.80. + 12½% V.A.T. p&p 35p.

AUDIO LEADS
FM Indoor Ribbon Aerial
3.5mm Jack plug to 3.5mm jack plug. Length 1.5m
5 pin DIN plug to 3.5mm. Jack connected
to pins 3&5. Length 1.5m
5 pin DIN plug to 3.5mm. Jack connected
to pins 1&4. Length 1.5m
Car aerial extension Screened insulated

115 5 pin DIN plug to 3.5mm. Jack connected to pins 18.4. Length 1.5m Car aerial extension. Screened insulated lead. Fitted plug & skt.

AC mains connecting lead for cassette recorders & radios. 2 metres
5 pin DIN phono plug to stereo headphone jack socket with attenuation network for stereo headphone jack socket with attenuation network for stereo districts of the plug to 18 most car cassette. B track cartridge & combination units. Supplied with inline fused power lead and instructions. 6.6m Coiled Guitar Lead Mono Jack Plug to 10 most car cassette. B track cartridge & combination units. Supplied with inline fused power lead and instructions. 6.6m Coiled Guitar Lead Mono Jack Plug to 5 pin DIN plug to 5 pin £0.85° 116 118 123

128

134 135

Please add 121% V.A.T. to all the above leads

DEPT. PE5, P.O. Box 6, Ware, Herts. Components Shop: 18 Baldock Street, Ware, Herts.

£0.60° €0 R5*

KITS FOR SYNTHESISERS, SOUND EFFECTS



COMPONENTS SETS Include all necessary resistors. capacitors, semiconductors, potentiometers and transformers. Hardware such as cases, sockets, knobs, keyboards, etc. are not included but most of these may be bought separately. Fuller details of kits, PCBs and parts are shown in our lists.

CIRCUIT AND LAYOUT OIAGRAMS are supplied free with all PCBs unless "as published"

PHOTOCOPIES of P.E. texts for most of the kits are available-prices in our lists.

PHONOSONIC

ORDER SUPPLIERS OF PRINTED CIRCUIT BOARDS, KITS AND TO WORLD-WIDE COMPONENTS A MARKET

P.E. MINISONIC Mk. 2 SYNTHESISER

A portable mains-operated Miniature Sound Synthesiser, with keyboard circuits. Although having slightly fewer facilities than the large P.E. Synthesiser the functions offered by this design give if great scope and versatility. Consists of 2 log VCOs. VCF. 2 envelope shapers, 2 voltage controlled amps, keyboard hold and control circuits. He oscillator and detector, ring modulator, noise generator, inter nowar supply. mixer, power supply.

Set of basic component kits (excl. KBD R's

nd tuning pots - see list for options available). from £61-00 Set of printed circuit boards

P.E. SYNTHESISER (P.E. Feb. 73 to Feb. 74)
The well acclaimed and highly versatile large-scale
mains-operated Sound Synthesiser complete with keyboard
circuits. Other circuits in our lists may be used with the Synthesiser to good advantage. Details in our lists.

FORMANT SYNTHESISER (Elektor 1977/78)

Very sophisticated music synthesiser for the advanced con-structor who puts performance before price. Details in our lists.

128-NOTE TUNE-PROGRAMMABLE SEQUENCER

Enables a voltage controlled synthesiser to automatically play pre-programmed tunes of up to 32 pitches and 128 notes long. Programs are keyboard initiated and note length and rhythmic pattern are externally variable. (Please use order codes quoted in

Main Circuit (Nov) excl. sw's (KIT 76-1)	£18-03
Power Supply (KIT 76-3)	£4.72
Trigger Inverter and Alt. Output (KIT 76-2)	£1-15
LED Counter (KIT 76-4)	£2-10
PCB (as published) for KITS 76-1 & 3 (PCB 76A)	£2-61
PCB for KITS 76-2 & 4 (PCB 76B)	£2.54

P.E. STRING ENSEMBLE (PE Mar-July 78)

The new keyboard string-instrument synthesiser.

fasic component sets:	
Power Supply (KIT 77-1)	£8-77
Tone Generator (KIT 77-2)	£14-66
Diode Gates (KIT 77-3)	£18-81
Chorus Generator (KIT 77-4)	£19-08
Voicing System (KIT 77-5)	£7-38
Printed Circuit Reade:	

Double-sided PCB for Power Supply, Tone Generator & Diode Gates with most of the Matrix wiring as printed tracking

(PCR //L/H)	£18.40
PCB for Chorus Generator (PCB 77C)	£2.65
PCB for Voicing System (PCB 77D)	£2.62
uller details of kits & PCRs are in our lists	

P.E. JOANNA PLUS ORGAN VOICING

P.E. JOANNA PLUS ORGAN VOICING
The basic five octave electronic piano (P.E. May/Sept 75 and
Sound Design) has switchable alternative voicings for HonkyTonk, ordinary piano, and Harpsichord or a mixture of any of these
three, together with facilities including fast and slow tremolo,
loud and soft pedal switching, and sustain pedal switching. The
modification retains all the circuitry associated with the piano
but in addition provides an organ-voice envelope facility with 5
switchable pitches, variable attack and sustain, phasing and
without

brato.

Set of components (excl switches) for PSU, Frequency generator, Pitch and Note Divider, Envelope Shapers, Voicings, and Control circuitries. (Order as KIT 71-5)

£99-25
Set of PCBs (Order as PCB SET 71-6)

£29-18

GUITAR EFFECTS PEDAL (P.E. July 75)
Modulates the attack, decay and lifter characteristics of an audio signal not only from a guilar but from any audio source, producing 8 different switchable effects that can be further modified by manual controls. Possibly the most interesting of all the low-priced sound effects units in our range. Circuit does not duplicate effects from the Guitar Overdrive Unit.

Component set with special foot operated switches Alternative component set with panel switches
Printed circuit board £5.05

FLEKTOR FLECTRONIC PLAND (Flektor Sept 78)

A touch-sensitive, multiple-voicing 5 octave piano using the latest integrated-circuit techniques for the keying and envelope shaping and virtually eliminating "be-hive" noise hitherto inherent in previous electronic planos. Details in our literature.

DIGITAL REVERBERATION UNIT (Elektor May 78)

A very advanced unit using sophisticated i.e. techniques instead of mechanical spring-lines. The basic delay range of 24 to 90mS can be extended up to 450mS using the extension unit. Further delays can be obtained using more extension.

Main component set (KIT 78-1)	£45.45
Extension component set (KIT 78-2)	£43-36
PCB for Kit 78-1 (PCB 78A)	£2.86
PC8 for Kit 78-2 (PCB 78B)	€1.0€

ANALOGUE REVERBERATION UNIT (Elektor Oct 78)

Using i.c.s Instead of spring-lines, the main unit has a maximum delay of up to 100mS, and the additional set extends this up to 200mS. May be used in either mono or Main component set (KIT 83-1)
Additional Delay Set (KIT 83-2)
PCB (as published) to hold both above kits (PCB 9973)

RESONANCE FILTER (Elektor Oct 78)

This filter module has been designed to allow a synthesis to produce a more realistic simulation of natural music

Basic component set (KIT 82-1)
PCB (as published) (PCB 9951) SYNTHESISER EXTERNAL INPUT INTERFACE

This unit allows external inputs, such as guitars, micro-phones etc. to be processed by the circuits within a synthesiser. Basic component set (incl PCB) (KIT 81-1) £2.94

GUITAR MULTIPROCESSOR (P.E. Dec/Feb 78)

An extremely versatile sound processing unit capable of producing, for example, Flanging, Vibrato, Reverb, Fuzz and Tremolo as well as other fascinating sounds. May be used with most electronic instruments. Details in our lists.

RHYTHM GENERATOR KITS

GUITAR FREQUENCY DOUBLER (P.E. Aug. 77)

A modified and extended version of the circuit publisi Component set and PCB

GUITAR SUSTAIN (P.E. Oct 77)

Maintains the natural attack whilst extending note duration.

Component set, PCB and foot switches £5.13 Component set, PCB and panel switches £3.71

WIND AND RAIN UNIT

A manually controlled unit for producing the above-named sounds Component set (incl. PCB)

GUITAR OVERDRIVE UNIT (P.E. Aug. 76)
Sophisticated, versatile Fuzz unit, including variable and switchable controls affecting the fuzz quality whilst retaining the attack and decay, and also providing filtering. Does not duplicate the effects from the Guitar Effects Pedal and can be used with it and with other electronic instruments.
Component set using dual slider pot Component set using dual rotary pot £6.89
Printed circuit board £1.62

Simple Fuzz unit based upon P.E. "Sound Design" circuit. Component set (incl. PCB)

TREMOLO UNIT £2.94

TREBLE BOOST UNIT (P.E. Apr. 76)
Gives a much shriller quality to audio signals fed through it.
The depth of boost is manually adjustable.
Component set (inct. PCB)
£2-51

WAVEFORM CONVERTER

Slightly modified from a circult published in "Elektor". Converts a saw-tooth waveform into four different waveforms; sine-wave, mark-space saw-tooth, regular triangle form, and squarewave with an externally variable mark-space ratio.

Component set (incl. PCB but excl. sw/s)

VOLTAGE CONTROLLED FILTER (P.E. Dec. 74)

Part of the P.E. Minisonic now released as an independent kit for use with other synthesisers.

Component set (incl. PCB) (Order as Kit 65-1) £7-17

RING MODULATOR (P.E. Jan. 75)
Part of the P.E. Minisonic now released as an independent kit for use with other synthesisers.
Component set (incl. PCB) (Order as Kit 59-1) £5-50

NOISE GENERATOR (P.E. Jan. 75)
Part of the P.E. Minisonic now released as an independent kit for use with other synthesisers.
Component set (incl. PCB) (Order as Kit 60-1)
£3-64

ENVELOPE SHAPER WITHOUT VCA (P.E. Oct. 75) Provides full manual control over attack, decay, sustain and release functions, and is for use with an existing voltage

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ENVELOPE SHAPER WITH VCA (P.E. Apr. 76)

This unit has its own voltage controlled amplifier and has full manual control over attack, decay, sustain and release functions.

Component set (incl. PCB)

TRANSIENT GENERATOR (P.E. Apr. 77)
An envelope shaper, without VCA, having the usual attack, decay, sustain and release functions, and in addition it also provides a "Repeat Effect" enabling a synthesiser to be programmed to limitate such instruments as a mandolin or habio.

Component set
Printed circuit board

SOPHISTICATED PHASING AND VIBRATO UNIT
A slightly modified version of the circuit published in
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A simple but effective manually controlled unit for introducing the "phasing" sound into live or recorded Component set (incl. PCB)

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VOICE OPERATED FADER (P.E. Dec. 73)

For automatically reducing music vo "talk-over"—particularly useful for Disco home-movie shows. Component set (incl. PCB) volume during or for

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AND OTHER PROJECTS

PHOTOGRAPHS in this advertisement show two of our units containing some of the P.E. projects built from our kits and PCBs. The cases were built by ourselves and are not for sale, though a small selection of other cases is available.

LIST—Send stamped addressed envelope with all U.K. requests for free list giving fuller details of PCBs, kits and other components.

OVERSEAS enquiries for list: Europe send 20p; other countries—send 50p.



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Type GA: 1 pair of contacts, normally open
Type GB: 2 pairs of contacts, each pair normally open
Type GC: 3 pairs of contacts, each pair normally open
Type GC: 4 pairs of contacts, each pair normally open
Type GE: 4 pairs of contacts, each pair normally open
Type GH: 5 pairs of contacts, each pair normally open each 251p each 24p each 281p

each 374p each 46 p Type 4PS: 3 pairs of contacts plus single-pole changeover each 57p

Printed Circuit Boards for use with most contacts (thus eliminating much interwining) are available. Details in our lists

P.E. TUNING FORK (P.E. Nov. 75)

Produces 84 switch-selected frequency-accurate tones. A LED monitor clearly displays all beat note adjustments. Ideal

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Main component set (incl. PCB)

Power supply set (incl. PCB) £6-28

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Component set
Printed circuit board

*This kit & PCB are at 8% VAT (all others are 12 \frac{1}{2}%) £24-05° £3.03°

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723 14-pin DIL	51p
726 TO5	1005p
741 8-pin DIL 748 8-pin DIL 4007 14-pin DIL 4011 14-pin DIL 4024 14-pin DIL 4069 14-pin DIL	24p
/48 8-pin UIL	57p
4007 14-pin DIL	17 p
4011 14-pin DIL	171p
4024 14-pin DIL	46 Jp
4069 14-pin DIL	18p
4 130 14-pin UIL	1200
AM2833 8-pin DIL	
AY10212 16-pin DI	
AY16721/6	188p
CA3046 14-pin DI	L /IP
CA3080 8-pin DIL	QCG
CA3084 14-pin DI	
FX209 16-pin DI	
LM323	562p
M252 16-pin DI	LOBUP
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AC176	28p
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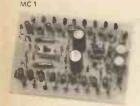
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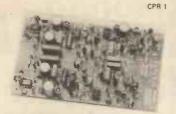
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XO2 – two way, XO3 – three way, Slope 24dB/Octave, Crossover points set to order within 10%.

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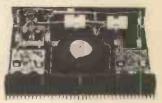
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W04		400V	0.40	PW06	6A	600V	1.18	S06	2A	600V	0.70
W06		600V	0.50	PW08	6A	800V	1.28	S08	2A		0.84
W08		8000	0.60	PW10	6A	10001		\$10	2A	1000	
VM18	1A	1001	0.44	K005		50V	2.20	B40C150C			0.53
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74C32N	0.24	74C107N	1.22	74C175N	0.90	#4C906N	0.54	74C926N	5.22
74C42N	0.92	74C150N	4.14	74C192N	1.11	74C907N	0.54	74C927N	5.22
74C48N	1.38	74C151N	247	74C193N		74C908N	n 96	74C928N	5.22
74C73N	0.54	74C154N		74C195N		74C909N		74C932N	
74C74N	0.56	74C157N		74C200N		74C910N		,	
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SN74H30N	0.55	74LS86N	0.44	74LS240N	1.50	SN74\$10N	0.77
SN74H40N	0.55	74LS90N	0.54	74LS241N	1.50	SN74S20N	0.77
SN74H51N	0.55	74LS91N	1.20	74LS242N	1.25	SN74S40N	0.77
SN74H53N		74LS92N	0.70	74LS243N	1.25	SN74SB4N	0.77
SN74H54N	0.55	74LS93N	0.64	74LS244N	1.50	SN74S65N	0.77
SN74H55N		74LS95AN	0.90	74LS245N	1.65	SN74S112N	1.70
SN74H60N		74LS96N	1.35	74LS247N	1.09	SN74S114N	1.70
SN74H62N		74LS107N	0.42	74LS248N	1.09	SN74S140N	0.77
SN74L00N	0.55	74L\$109N	0.42	74LS249N	1.09	SN74S157N	2.95
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74LS20N	0.26	74LS157N	0.60	74LS325N	2.40	SN7493N	0.36
74LS21N	0 26	74LS158N	0.65	74LS326N	2.70	SN7494N	0.90
74LS22N	0.26	74LS160N	0.80	74LS327N	2.55	SN 7495N	0.76
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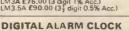
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		3.35		2.50		0.75	LF355N	
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2N335 3.00		0.30	2N1420 0.55	2N2219 0.38	40081	1.20	AC188	0.54	BC118	0.22	ВСҮ33А	1.10	B0140	0.43	BF157	0.48
2N388 0.80		0.30	2NE483 1.85	2N2219A 0.39	40232	0.60	AC188K	0.65	BC119	0.33	BCY34A	1.10	BD142	0.70	BF159	0.37
2N388A 0.77		0.54	2N1485 2.20	2N2220 0.39	40233	0.70	AD136	2.75	BC122	0.61	BCY38	2.20	BDX14	1.32	BF160	0.33
2N398 0.8D		0.85	2N1507 0.35	2N2221 0.25	40235	0.65	AD142	1.45	BC123	0.65	BCY40	1.05	BDX18	1.90	BF161	0.65
2N456 2.20		1.05	2N1524 0.60	2N2221A 0.25	40237	0.45	AD143	1.45	BC125	0.22	BCY42	0.65	BDX32	2.67	BF 166	0.44
2N489 4.90		0.45	2N1553 1.50	2N2222 0.25	40242	0.66	AD149	2.85	BC132	0.33	BCY43	0.76	BDY11	3.30	BF167	0.37
2N489A 5.40		0.50	2N1613 0.30 2N1637 0.72	2N2222A 0.25 2N2223 -6.78	40251	0.66	AD150 AD161	3.10	BC134	0.22	BCY54	2.40	BDY17	3.30	BF170	0.76
2N489B 5.90 2N490 4.99		0.35	2N1637 0.72 2N1638 0.70	2N2223 A 6.99	40254	0.95	AD162	1.00	BC135 BC136	0.22	BCY58 BCY59	0.27	BDY18	3.80	BF173	0.37
2N4908 6.50		0.35	2N1711 0.30	2N2270 0.49	40280	3.70	AF306	0.60	8C137	0.21	BCY66	0.27 2.20	BDY20 BDY23	1.10	TIP29C	0.65
2N490C 6.9D		0.35	2N1889 0.30	2N23D3 1.54	40309	0.60	AF109	0.87	BC138	0.44	8CY67	2.70	BDY24	2.45	TIP30A TIP30C	0.54
2N491A 5.75		0.38	2N1890 0.30	2N23B8 0.27	40310	0.85	AF114	0.70	8C140	0.30	BCY70	0.21	BDY25	2.85	TIP31A	0.70
2N491B 6.25		0.33	2N1893 0.30	2N2369 D.27	40311	0.55	AF115	0.70	BC141	0.32	BCY71	0.26	BDY38	1.10	TIP31A	0.72
2N491 7.50		0.38	2N1907 5.95	2N2369A 0.27	40312	0.99	AF116	0.70	BC141	0.32	BCY72	0.18	8DY54	1.55	TIP32A	0.72
2N492 5.25		0.45	2N1974 0.98	2N2405 0.66	40313	1.38	AF117	0.70	BC143	0.32	BCY77	0.70	BDY55	1.90	TIP32C	0.82
2N492A 6.75		0.37	2N1990 0.45	3N81 3.50	40315	0.60	AF118	0.70	BC1 52	0.38	8CY78	0.43	BDY56	2.10	TIP33A	0.86
2N492B 7.75		0.37	2N1991 1.10	3N83 2.25	40316	0.95	AF124	0.70	BC153	0.30	BCY79	0.41	BDY57	5.90	TIP33C	1.18
2N492C 10.00		0.37	2N2060 7.00	3N128 1.35	40317	0.60	AF125	0.70	BC154	0.30	BCY87	5.35	BDY58	6.50	TIP34A	0.97
2N493A 7.99	2N93DA	0.95	2N2102 050	3N139 1.60	40324	0.95	AF126	0.70	BC160	0.38	8CY88	3.99	BDY60	1.65	TIP34C	1.31
2N493B 8.75	2N1131	0.32	2N2147 1.55	3N140 1.10	40325	1.35	AF139	0.75	BC161	0.38	BCY89	3.80	BDY61	2.75	TIP35A	2.20
2N494 6.90	2N1132	0.35	2N2160 1.55	3N141 0.95	AC126	0.48	AF172	0.70	BC175	0.43	BD115	88.0	BDY62	2.75	TIP36A	3.00
2N494A 7.65	2N1204	1.65	2N2192 0.58	3N142 0.70	AC127	0.48	AF178	1.30	BC204	0.12	BD116	1.35	8DY92	2.75	TIP41A	0.76
2N494B 8.40	2N1302	0.B0	2N2193 0.5D	3N143 0.88	AC1 28	0.48	AF186	0.55	8C205	0.17	BD121	2.20	BF115	0.39	TIP41C	0.97
2N494C 9.35		0.80	2N2193A 0.52	3N152 1.10	AC151	0.43	AF200	1.30	BC206	0.17	BD124	2.20	BF119	1.10	TIP42A	0.86
2N549 3.25		08.0	2N2194 D.42	3N153 1.89	AC152	0.54	AF201	1.30	BC207	0.17	BD131	0.55	BF121	0.60	TIP42C	1.08
2N696 0.39		0.80	2N2194A D.45	3N154 0.99	AC153	0.59	BC113	0.22	BC208	0.17	BD132	0.75	BF123	0.60	TIP50	0.82
2N697 0.31		1.00	2N2195 0.49	3N159 1.35	AC153K	0.59	BC114	0.22	BC209	0.17	BD135	0.40	BF134	0.60	TIP54	1.83
2N698 0.49		1.00	2N2195A 0.49	3N187 1.B0	AC176K	0.70	BC115	0.22	BCY10	1.10	BD136	0.40	BF137	0.60	TIP110	0.77
2N699 0.5B		1.10	2N2217 0.55	3N200 2.85	AC176	0.54	BC116	0.21	BCY30A	1.10	BD137	0.41	BF152	0.27	TIP112	0.93
2N706 0.30		1.10	2N2218 0.35	3N201 1.35	AC187	0.59	JBC116A	0.22	BCY31A	1.10	BD136	0.41	BF153	0.27	TIP115	0.83
2N706A 0.30	2N1370	0.55	2N2218A 0.38	40050 1.70	AC187K	0.65	BC117	0.22	BCY32A	1.10	BD139	0.43	BF154	0.27	TIP117	0.99

Practical Electronics May 1979

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FROM time to time we see large industrial distributors or manufacturers breaking into the component retail market; this has happened again just recently. They obviously feel the market is a lucrative one; they certainly enter it with both eyes open and usually after considerable research—sometimes with quite a fanfare too! Unfortunately, nearly all these companies have in the past come to grief, having struggled along for a few years making little or no profit and finally giving up and closing down.

We were fortunate enough to be invited to the press launch for Doram-it was an impressive affair in a wine cellar in the city of London. The link between the fledgling company and Radiospares—as RS Components were then known-was emphasised. There was a display of kits and components that Doram would market and directors and technical staff were on hand to answer the many questions from the hobbyist and general electronic press. Those who doubted the viability of the project, in the light of the failure of other companies, were soon guietened by the enthusiasm and professionalism of the Doram people.

Alas it was not to be, and having struggled along for a few years, a recent change in policy to get out of components and just sell kits was eventually squashed by the decision to shut up shop completely. How then can so many "private" firms continue.

Perhaps it is simply that "industrial" companies need a bigger profit margin? Maybe they use over complex systems, designed for industry and not mail order? Or could it simply be that their marketing is all wrong for the hobbyist? It's probably a combination of all these factors. One thing is clear, our market is not an easy one to judge, or to be profitable in. Of course, we wish all "new" companies well and we hope the latest venture by a distributor will be successful—only time will tell.

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Many successful component suppliers advertising in this magazine have started with one or two people, often from someone's front room or a small back street office. They have built up a business from a spare time occupation and now many of them employ upwards of a dozen people. Perhaps because they started with very little capital they have learnt how to keep the overheads down, how to market to

you the hobbyist consumer, and above all how to work and make a profit on a low margin.

We do, of course, get some complaints about suppliers, long delivery times, etc. but they are often faced with no delivery from manufacturers and the breaking of promised delivery dates from the world's suppliers. There is the odd bad company, but rest assured they are quickly weeded out of our pages and our mail order protection scheme is used to help those who's money has gone to a bankrupt company.

By and large we have nothing but praise and admiration for the component suppliers; most of them operate a fast, efficient service at an excellent price. They carry many thousands of different items with a large proportion available for pence rather than £s. It is necessary for them to buy stock directly from many countries, to rely on the post for most of their sales and to operate in a highly competitive field. Industry has shown many times that it is willing, but often unable to operate successfully in this area, so without the private retail component suppliers our hobby would probably cease, certainly be more expensive.

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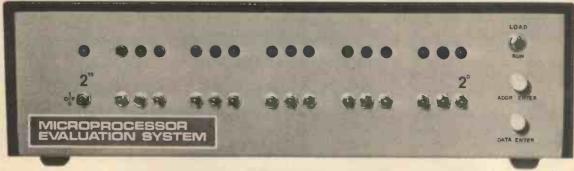
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Practical Electronics May 1979

Microprocessor Evaluation System

D.S.COUTTS (General Instrument Microelectronics)



Part 1

WITH microprocessors steadily spreading through the electronics hobby scene, the author felt that it was time to learn a little about them and how they can be programmed to perform simple tasks. The question was, where to start? The requirement was for a very simple machine using a powerful processor so that the lessons learned could be applied later to build a more sophisticated machine. The simplest approach seemed to be to use switches to read in the address and data information (eliminating the need for a PROM programmer) and to use single l.e.d.s as the output display. The microprocessor chosen was the GENERAL INSTRUMENT CP1610, a 16 bit machine with a very powerful instruction set. A large memory wasn't considered necessary to learn the rudiments

PSU

-3V +5V +1IV

-3V 0V +5V +1IV

CLOCK
RESET CIRCUIT

CP1610
C PU

CP1610
C PU

CONTROL
CIRCUIT

ADDRESS
LATCH (8 BITS)

LATCH (8 BITS)

Fig. 1.1. Block diagram of the Evaluation System

of programming so four 2112 RAMs were used, giving a total of 256 sixteen bit words (1,024 bits).

The basic system block diagram is shown in Fig. 1.1. It consists of a power supply, the CP1610 CPU, TTL clock, reset and control circuits, four 2112 RAMs and TTL address and output latches. Data and addresses are set up by 16 s.p.d.t. toggle switches each with a centre off position. There is also a LOAD/RUN toggle switch and two push buttons for ADDRESS ENTRY and DATA ENTRY. 16 single l.e.d.s are used for the output. The memory occupies address space \mathscr{B} –377 OCTAL (\mathscr{B} –255 BINARY) and the output latches are at address 377 OCTAL (255 BINARY).

INTRODUCTION TO CP1610

The CP1610 internal block diagram is shown in Fig. 1.2. There is an array of 8 sixteen bit registers. Register 6 serves as the stack pointer (SP) while register 7 serves as the program counter (PC). The other registers are general purpose and may be used for arithmetic or other operations.

Communication to the outside world is via a 16 bit bidirectional bus, but internally the data is processed in two 8-bit bytes. There are four status flags SIGN (S), ZERO (Z), OVERFLOW (OV) and CARRY (C), and there is a 10 bit instruction register. Table 1 shows the CPU pin connections and functions.

All the pins are not used in our basic processor, pins 27, 28, 31, 32, 33 are linked to +5V and pin 40 also goes via a resistor to +5V. Pins 1, 22, 23, 24, 25 are not used at first. This leaves us with four power input pins, two clock input pins, a reset input, sixteen I/O lines for data and address and three output pins to decode for bus control purposes. Not too frightening!

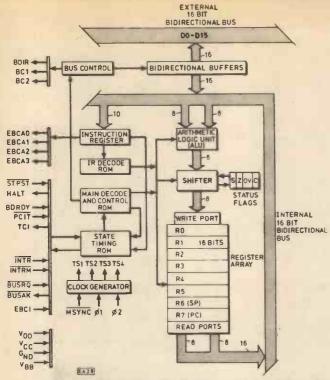


Fig. 1.2. Internal block diagram of the CP1610

The processor instruction set is shown in Table 2. It may seem a bit overpowering at first but there is a glossary of terms at the end of the table and it is intended to give programming examples when the processor has been completed.

CIRCUIT DESCRIPTION

The complete circuit diagram of the system is shown in Fig. 1.3. The circuit is relatively straightforward and will be described in blocks.

CLOCK CIRCUIT

IC1 a, b and c are used for the master oscillator and this feeds the shift register 2a, b and 3a, b. This shift register is used to produce four time slots, TS1 to TS4. IC5a and b are used to produce two non-overlapping clocks \$\mathscr{\textit{9}}1\$ and \$\mathscr{\textit{9}}2\$. These clocks are buffered by the open collector TTL IC6a and b, and drive the CPU via emitter followers TR1 and TR2.

LOAD/RUN AND RESET CIRCUIT

When S1a is in the load position IC4 pip 8 is low, preventing the shift register from running and stopping the clocks to the processor. The low on IC4 pin 8 is fed to IC4 pin 5, holding pins 6 and 2 high. As pin 1 is also high, pin 3 is held low, holding the CPU in the reset state. In this state the sixteen processor I/O lines are in their high impedance state and the switches (S4 to S19) may be used to load the program into RAM. When S1a is changed to the run position the CPU clocks are enabled and the CPU reset line goes high when IC3a pin 9 goes low, and the program is executed.

BUS CONTROL

IC8, a B.C.D. (decimal decoder) is used to decode the BC1, BC2 and BDIR lines from the processor to produce the necessary bus control signals. BC1 is on pin 3, BC2 on 4 and BDIR on 5. Pin 12 is fed from IC5c and goes high during reset and also during time slot TS1. This prevents IC8 from outputting unwanted signals during reset and also during

VCC VBB VDD D0-D15 MSYNC	, + C	-5V -3V -11V ATA AN		RESS LINES	
EBCAØ EBCA1 EBCA2 EBCA3		bits of i to sele sample	SS. Out	puts of 4 least s in register, used 16 digital stat e CPU during	externally es to be
EBCI BDRDY	Inp		L BRAN		
INTR INTRM TCI BUSRQ	Ou	tput, te	rminates	e maskable current interrupt U to relinquish	
BUSAK STPST HALT PCIT	Ot Alt	itput to ternately itput, inc	y stops a dicating	at CPU has releand starts CPU. CPU halted. vents the progra	
		increme instruct As out	enting tions. put, outp	during fetch pouts negative punterrupt) instruct	ohase of
BC1 BC2 BDIR	В	is Controls Controls DIRec	ol 1 ol 2	see bus control signals below	
BDIR I	BC1	BC2		DECODE FUN	CTION
0	0	0	NACT	No ACTion,	
				D0-D15=high	1 - 1
	_			impedance	
0	0	1	IAB	Interrupt Address Bus, D0-D15=	ss to
0	1	0	ADAR	Address Registe	to er, D 0
0	1	1	DTB	D15=high impe	
1	0	0	BAR	=Input Bus to Address	
1	0	1	DWS	Register Data Write	D0-D15
1	1	0	DW INTAK	Strobe Data Write INTerrupt	=Output
BC1 C BC2 D BCIR C	2 3 4 8 8 7 8 9 9 110 111 12 13	40 PCIT 30 QMO 30 0 01 30 0 01 30 0 02 30 0 700 (*) 35 0 700 (*) 35 0 80R0 32 0 \$TPST 31 0 BUSRC 30 1 HALT 22 0 BUSRC 26 0 INTR 27 0 INTRA 28 0 TCI	TAPO TAPO TAPO TAPO TAPO TAPO TAPO TAPO	AcKnowledge BLE 1. Complete croprocessor to the croprocessor to t	1610 ogether

26 D TCI 25 D EBCAO 24 D EBCAO 23 D EBCA2 22 D EBCA3 21 D O2

REGISTER	REGISTER-REGISTER	ER					
MNEMONIC	OPERAND	CYCLES	SNI	INSTRUCTION		DESCRIPTION	STATUS CHANGE
MOVR	SSS, DDD	.9	0100	SSS	000	MOVe contents of Register SSS to register DDD. *H DDD is	5, 7
TSTR	555	. 9	0000	555	SSS	TeST contents of Register SSS. "H SSS is B or 7 add	5, 2
J.R	SSS	1.	0100	SSS	Ξ	Jump to address in Register SSS. (Move address to Register 7)	5, 2
ADDR	888, 000		1100	\$55	000	ADD contents of Register SSS to contents of register DDD. Results to DDD	S, Z, C, DV
SUBR	SSS, DDD	60	0100	255	. 000	SUBtract of Register SSS from contents of register DDD. Results to DDD.	S, Z, C, 0V
СМРЯ	SSS, DDD	80	1010	255	000	CoMPare Register SSS with register DDD by subtraction. Results not stored.	S. Z. C. 0V
ANDR	SSS, DDD	٤	0110	SSS	000	logical AND contents of Register SSS with contents of register DDD. Results to DDD.	5, 2
XORR	888, 000	60	1110	SSS	000	eXclusive OR contents of Register SSS with contents of register 000. Results to 000.	2.2
CLRR	000	60 66	0111	000	000	CLeaR Register to zero. NCrement contents of Register DDO. Results to DDO.	5,7
DECR	000	9	0000	010	000	DECrement contents of Register DDD, Results to DDD.	5, 7
COMR	000	(D)	0000	110	000	One's COMplement contents of Register DDD. Results to DDD. Two a complement contents of Register DDD Results to DDD.	5, Z
ADCR	000	p co	0000	101	000	ADd Carry bit to contents of Register DDD. Results to DDD.	S, Z, C, OV

		_	_		_	_					
	5,2	5, 2	5, 2	S, Z, C, 0V	S. Z. C	S. Z. C. 0V	5,2	2.3	S. Z. C. 0V	S, Z, C	S. Z. C. 0V
Executable only with Register 0, 1, 2, 3, n=1 or 2 (places) Shift Right instructions set the S flip-flop with Bit 7 of the result after the Instruction. Add 2 cycles if Bhifti is 2 bit or two bytes. Shifts are not interruptable.	N=0, SWAP bytes of register RR. S equals Bit 7 of results of SWAP.	N=1, SWAP bytes of register RR, then swap them back to onginal form	N=0. Shirk Logical Left one bit, zero to low bit. N=1, Shirk Logical Left two bits, zero to low 2 bits.	N=0, Rotate Left one bit using Carry bit as bit 18. N=1, Rotate Left two bits using C as bit 17 and OV as bit 16	N=0, Shift Logical Left one bit using C as bit 16,	N=1, Shift Legical Left two bits using C as bit 17, 0V as bit 16, zero to low 2 bits	N=0. Shift Logical Right one bit, zero to high bit N=1. Shift Logical Right two bits, zero to high two bits.	N=0, Shift Anthmetic Right one bit, sign bit copied to high bit, N=1, Shift Arithmetic Right two bits, sign bit copied to	high bits. N=0. Rotate Right one bit using Carry as bit 16. N=1, Rotate Right two bits using C as bit 16. 0V as bit 17.	N=0. Shirt Arithmetic Right one bit, thru Carry, sign bit cooled to high bit.	M=1. Shift Arithmetic Right two bits, thru Carry and OV, sign bit copied to high 2 bits
flop wit	NAR		NRR	NAR	NRR		NRR	N-RR	NRR	NRR	
r 0, 1, 2, ne S flip or two b	000		100	010	011		100	101	110	Ξ	
Registe ons set th is 2 bits o	1000		1000	1000	1000		1000	1000	0001	1000	130
REGISTER SHIFT Executable only with Register 0, 1, 2, 3, m=1 or 2 (piaces) Shift Right instructions set the 5 flip-flop with Bit 7 of th Add 2 cycles if shift is 2 bits or two bytes. Shifts are not interruptable.	. 60	80	မွာ	10 KD, 0	0 40	00	ထင	10 00 00	. 60 0	9	8
SHIFT EX	RR<,n>	,	RR<,n>	RR<,n>	RR<,n>		RR<,n>	RR<,n>	RR<.n>	RR<,n>	
REGISTER	SWAP		1118	RIC	3118		SLR	SAR	RRC	SARC	

ISTER	ISTER SHIFT EX Sh Ad	Executable only with Register 0, 1, 2, 3. n=1 or 2 (piaces) Shift Right instructions set the S flip-flop with Bit 7 of th Add 2 cycles If shift is 2 bits or two bytes. Shifts are not interruptable.	h Registe ons set th is 2 bits o	r 0, 1, 2, le S flip- ir two by	3. n=1 cflop with	Executable only with Register 0, 1, 2, 3 n=1 or 2 (places) Add 2 cycles if shift is 2 bits or two bytes. Shifts are not interruptable.		80E 8GT	DA DA
VAP	RR<,n>	. 9	1000	000	NAR	N=0. SWAP bytes of register RR. S equals Bit 7 of results of SWAP.		BUSC	DA
	,	80				N=1, SWAP bytes of register RR, then swap them back to original form		BESC	DA
	RR<,n>	90	1000	100	NRR	N=0. Shirk Logical Left one bit, zero to low bit. N=1. Shirk Logical Left two bits, zero to low 2 bits.		BEXT	DA. E
ບ	RR<,n>	na ko, t	1000	010	NAR	N=0. Rotate Left one bit using Carry bit as bit 18. N=1. Rotate Left two bits using C as bit 17 and 0V as bit 16. S. Z. C. DV			
31	RR<,n>	no 40	1000	110	NRR	_			
		00				N=1, Shirt Legical Left two bits using C as bit 17, 0V as S. Z. C. OV his figures had 2 his	_		
ÓT.	RR<_n>	\$ 0 (1000	100	NRR	e bit, zero to high bit obits.			
ac.	RR<,n>	30 OC OC	1000	101	NRR NRR	N=0. Shirt Anthmetic Right one bit, sign bit copied to high bit. S. Z N=1. Shirt Anthmetic Right two bits, sign bit copied to S. Z			
c)	RR<.n>	800	0001	110	NRR	high bits. N=0, readen Right one bit using Carry as bir 18. S. Z. C. N=1. Rotate Right two bits using Cas bir 16, 0V as bit 17. S. Z. C. 0V		JUMP	
RC	RR<,n>	9	1000	Ξ	N N N N N N N N N N N N N N N N N N N	N=0. Shift Arithmetic Right one bit, thru Carry, sign bit copied to high bit. M=1 Khit Anhmetic Right non bits than Carry and DV C 7 C DV		MNEMONIC	OPERAND
		ω	S					7	DA
								4	DA
VTROL								9	DA

			JUMP

Branch if External condition is True. Field E is externally decoded to select 1 of 18 conditions. Response is tested for true condition.

PP

pppp

Branch if Less than or Equal, ZV (S-V-0V)=1 Branch if Greater than or Equal, S-4-0V=0

dddddd dødødd

dddddd

8/7 8/7 8/7 8/7 8/7 8/7

Branch on ZEro. Z=1
Branch if EQual. Z=1
Branch on No ZEro. Z=0
Branch if Not EQual. Z=0
Branch if Not EQual. Z=0

Branch if Unequal Sign and Carry C.W.S=1 Branch if Equal Sign and Carry C-4-S=0

Branch Il Greater Than. ZV (S-V-0V)=0

DA 12 0.000 100 0.000 0.000 100 0.000 0.000 100 0.000	JUMP				4	
DA 12 0000 000 100 100 100 100 100 100 100	MNEMONIC	OPERAND	CYCLES	INSTRUCTION	DESCRIPTION	STATUS
DA 12 AAAA AAA AAA AAA AAA AAAA AAAA AAA		VC	12	1 000	himo to addrese. Program counter is set to 16 hits of A's	
DA 12 0000 000 100 100 100 100 100 100 100	2	Š	4	AAA		
DA 12 0000 0000 100 DA 12 0000 0000 100 DA AAAA AAA AAA BB. DA 12 0000 000 100 BB. DA AAAA AAA AAA AAAA AAAA AAAA AAAA			l	AAA		
11	当	DA	12	000	Jump to address. Enable interrupt system. Program counter is	
DA 12 AAAA AAA AAA AAA AAA AAA AAA AAA AAA				AAA	set to 16 bits of A's.	
12 0000 0000 100 1				AAA		
11 A A A A A A A A A A A A A A A A A A	20	DA	12	000	Jump to address Disable interrupt system Program counter	
8B. Da 12 0,000 000 100 100 88. Da 24A AAA AAA AAA AAA AAA AAA AAA AAA AAA				AAA	is set to 18 bits of A's.	
8B. DA 12 0000 000 100 8B.A.A.A.A.A.A.A.A.A.A.A.A.A.A.A.A.A.A.A				AAA		
88.0 A A00 A00 A00 A00 A00 B8.0 A A00 B8.0 A A00 A00 A00 A00 A00 A00 A00 A00 A00	JSR	BB, DA	12	000	Jump and Save Return address (PC+3) in register designated	
88. DA 12 0006 000 100 100 88. DA AAA AAA AAA AAA AAA AAA AAA AAA AAA				AAA	by 1BB. Program counter is set to 16 bits of A's BB#11	
8B. DA 12 0000 0000 100 100 100 100 100 100 10				AAA		
88.0A 12 0000 000 100 88.0A AAA AAA AAA AAA AAA AAA AAA AAA AAA	JSRE	BB, DA	12	000	Jump and Save Return and Enable interrupt system, Return	
88. DA 12 0000 000 100 88AA AAA AAA AAA AAA AAA AAA AAA AAA				AAA	(PC+3) Is saved in register 188. Program counter is set	
88.0a 12 0000 000 100 88AA AAA A10 AAAA AAA AAA				AAA	to 18 bits of A's. BB=11	
88AA AAA A10 AAAA AAA AAA	JSRD	88, DA	12	000	Jump and Save Return and Disable interrupt system. Return	
AAA AAA		1		AAA	(PC+3) is saved in register 18B. Program counter is set	
				AAA	to 16 bits of A's. BB=11	

SSWD DO SSTALUS WORD in register DO. Bits 40–3, 8–11 set to 0	CONTROL							
Correction Cor								
San	COMU	5	40	0000	110	000	Get Status WorD in register DD. Bits 0-3, 8-11 set to 0	
Carry Carr	2	00					Bits 4, 12=C; 5, 13=0V; 8, 14=2; 7, 15=5.	
Carry 10 Carry 10	NOP		9	0000	110	10N	No operation	
SSS 6 0000 111 SSS Retroat State Word from registry SSS-, Bit 4 to C.	SIN	(N)	99	0000	110	11N	Software Interrupt; pulse to PCIT " pin, n=0 or 1	
4 0000 000	RSWD	888	9	0000	111	SSS	Restora Status Word from register SSS; Bit 4 to C.	S. Z. C. 0V
4 0000 000 000 Half Their security Security on Security State Control Start. 4 0000 000 011 Disable Interrupt System. Not Interruptable 0000 011 Disable Interrupt System. Not Interruptable 1000 000 111 Terminals Control Interruptable 1000 000 110 Clask Earry to see, Min Interruptable 1000 000 111 SET Carry to one. Not Interruptable.		200				į	Bit 5 to 0V, Bit 6 to Z, Bit 7 to S.	
4 0000 010 010 Grable Internal System. Not Internationals. 4 0000 000 011 Grable Internal System. Not Internationals. 4 0000 000 101 Terminate Current Internationals. 4 0000 000 111 SET Carry to one. Not Internationals. 4 0000 011 SET Carry to one. Not Internationals.	1		4	0000	000	000	HaLT after next instruction is executed. Resume on control start.	
4 0000 000 101 Disable interrupt System, Mel Interruptable. 4 0000 000 101 Terminate Carry Interruptable. 4 0000 000 110 CLes Carry to one, Mel Interruptable. 4 0000 111 SET Carry to one, Mel Interruptable.	FIS		4	0000	000	010	Enable Interrupt System, Not Interruptable.	
4 0000 000 101 Terminate Curring New York (Not Interruptable. 2000 000 110 SET Carry to one. Not Interruptable. 4 0000 000 111 SET Carry to one. Not Interruptable.	DIC		4	0000	000	011	Disable Interrupt System, Not Interruptable	
4 0000 000 110 CLeaR Carry to zero. Not Interruptable. 4 0000 000 111 SET Carry to one. Not Interruptable.	TLI		4	0000	000	101	Terminate Current Interrupt. Not Interruptable.	
4 0000 000 111 SEI Carry to one. Not Interruptable.	7017		4	0000	000	110	CleaR Carry to zero. Not Interruptable.	٢
	CETT		4	0000	000	Ξ	SET Carry to one. Not interruptable.	ں
	2000							
						Ī		

The Branch Instructions are Program Counter Relative, i.e., the Effective Address=PC_EDisplacement, PPPPPPPPP is the Displacement and Si at 0 for + 1 for - 1 ff Mamory is greater than 10 bits then the appropriate number of lead its papepp will be a part of the Displacement. For sforward branch and addition is performed; for a backward branch a ones complement subtraction is performed. Computation performed on PC+2.

ANCHES

STATUS

DESCRIPTION

INSTRUCTION

CYCLES

OPERAND DA

EMONIC

1/8

NOPP

1/8 1/9 1/9 1/8

BLGE BNC BNC BLLT BDV BNOV

Branch unconditional, Program Counter Relative (+1025 to —1024)
No OPeration, two words

Branch on Carry, C=1
Branch it Logical Greater Than, or Equal, C=1
Branch on No Carry, C=0
Branch it Logical test Than, C=0
Branch to OVerflow, OV=1

Branch on No OVerflow, OV=0

Branch on Minus S=1 Branch on Plus S=0

10000 10000

BPL BMI BED BED BNED BNED BNED

1/9 7/9

DIRECT ADDRESSED DATA—MEMORY Field age age is dependent on the width of	ODRESSEL to is depend	DATA_A	MEMOR'	rf memor	ory. 16 bi	ts is ma	DIRECT ADDRESSED DATA—MEMORY Field ase ase is dependent on the width of memory. 16 bits is maximum for ase ase AAAAAAAAAA.	
MNEMONIC	DPERAND	CYCLES		INSTR	INSTRUCTION		DESCRIPTION	STATUS
MVD	SSS, A	11		1001	000	SSS	MoV Out data from register SSS to address A-A.	
MVI	A. 000	. 01	333 333	1010	ODO ODO	900	MoVe In data from address AA to register DDB.	
ADD	A, 000	1	868 868 4	1011	000	AAA	ADD data from address A-A to register DDD. Results to DDD.	S.Z.C. 0V
SUB	A. DDD	ç	888 888	1100	AAA	AAA	SUBtract data from addiess A A from register DDD Becults	S 7 E DV
dMJ.	V 666	2 5	899 999	AAAA	AAA	AA S	to DOD.	0 7 F DV
AND	200	2 9	888 888	AAAA	AAA	\$ \$ 50 5 4 50 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5	Subtraction. Results not stored.	3, 2, 0
XOR	A, DOD	2 2	883 338	AAAA 1111	900 000	98 e	Results to DDD. eXclusive DR data from address A—A with register DDD.	5. Z
			898 999	AAAA	AAA	AAA	Results to DDD.	

S. Z. C. 0V

Move in double byte data from the address in register MMM to register MMM to the south of control of register DDC.

ADD double byte data from the address in register MMM to the control of register DDC.

SUBINACI double byte data located as address MMM from the control of register DDD. Results to DDD.

Could are double byte data located as address in register MMM swith the control of register SDS by subtraction. Results with the control of register SDS by subtraction. Results with the control of register SDS by subtraction. Results MMM with the control of register DDD. Results to DDD.

Exclusive OR double byte data located as address in register.

MMM with the control of register DDD. Results to DDD.

Exclusive OR double byte data located as address in register.

0000 11010 0000 1100 1101 1101 0000 00000 11110

000 MMM 000 MMM

4040

000 MMM 000 MMM 000 MMM 000

40404040

MMM, DDD MMM, 000

MMM, DDD

SDBD MVI® SOBD SDBD SDBD CMP® SDBD CMP® SDBD CMP® SDBD AND® SDBD AND® SDBD XOR®

MMM, DDD MMM, DD0 MMM, DDD

5.2

STATUS

DESCRIPTION

INSTRUCTION

CYCLES

OPERAND

MNEMONIC

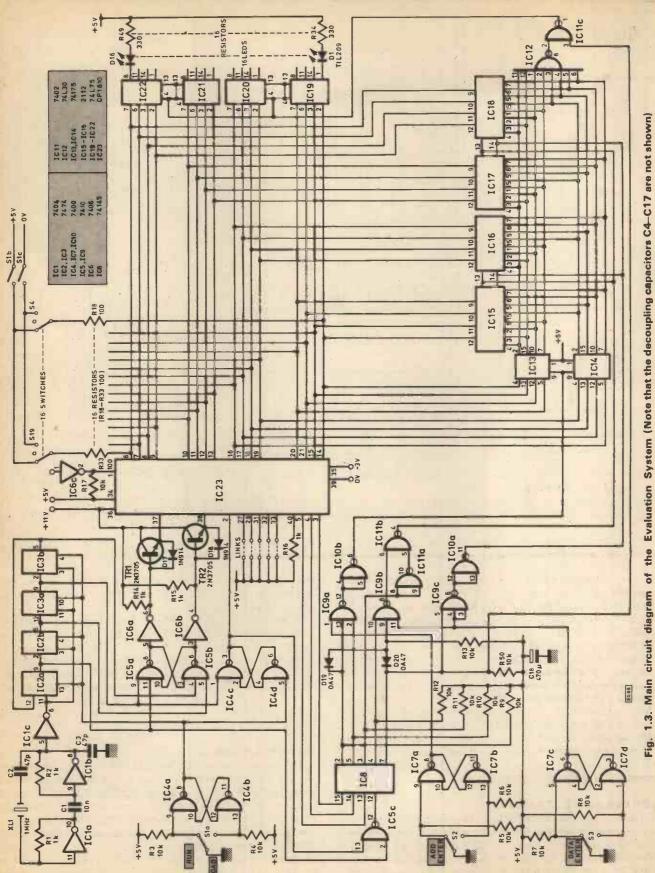
INDIRECT ADDRESSED DOUBLE BYTE DATA-REGISTER

						MoVe Out data from register SSS to the address in register	
MVO@ SS	SSS. MMM	8	1001	MMM	SSS		
						MMM. Note: SSS=MMM=4, 5, 6 or 7 not supported.	
Ī	SSS	o 1	1001	011	SSS	PuSH data from Register SSS to the stack.	
MVIG	MMM. DDD	. :	0101	MMM	000	MoVe in date to register DDD from address in register MMM.	
Ī	nnn	= 1	0101	011	000	-	
ADD(6 MI)	MMM, DDO		1011	MMM	000	ADD data located at address in register MMM to the contents	S. Z. L. UV
		,				of register DDD. Results to DDD.	
SUB(MMM, DDD	000	1100	MMM	000	MMM from	S. Z. C. 0V
Ī				-			
CMP (a MA	MMM. DDD		1101	MMM	SSS	CoMPare data located at address in Register MMM with	S. Z. C. UV
Ī						contents of register SSS, by subtraction. Results not stored	
AND(s	MMM, DDD	000	1110	MMM	000	logical AND contents of register DDD with data located at	S. Z
Ī						address in register MMM. Results to 000.	
XOR(6) MA	MMM, DDD	00	1111	MMM	000	eXclusive OR contents of register DDD with data located at	2.2
						address in register MMM. Results to 000.	

						S. Z. C. 0V		S. Z. C. 0V		S. Z. C. 0V		2.2		2.2	1	
LODRESSED DATA—REGISTER redats in contract at the address contained in Register, R1—R6. WO Instruction—post increment R6 PUSM data from Register SSS to the Stark Dither instructions—pre decrement R6 PUSM data from the Stark to be used as the first operand.	MoVe Out data from register SSS to the address in register	MMM. Note: SSS=MMM=4, 5, 6 or 7 not supported.	PuSH data from Register SSS to the stack.	MoVe in date to register DDD from address in register MMM.	PULI data from the stack to Register DDD.	ADD data located at address in register MMM to the contents	of register DDD. Results to DDD.	SUBtract data located at address in Register MMM from	contents of register DOD. Results to DDD.	CoMPare data located at address in Register MMM with	contents of register SSS, by subtraction. Results not stored	logical AND contents of register DDD with data located at	address in register MMM. Results to DDD.	eXclusive OR contents of register DDD with data located at	address in register MMM. Results to 000.	
11—RG. Register SSS to the Stack to be	SSS		888	000	000	000		000		SSS		000		000		
R n Register. F I data from L data from	MMM		110	MMM	110	MMM		MMM		MMM		MMM		MMM	-	
REGISTE is contained i lent R6 PUS) ment R6. PUS	1001		1001	1010	1010	101		1100		1101		1110		1111	-	
CTADDRESSED DATA—RÉGISTE R Source data in courte at the address contained in Register. R1—R6. 4. 5 poir increment R4 or R5. 6—MVO Instruction—post increment R6 PUSM data from Register SSS to the Stack Dithe instructions—pre decrement R6 PULL data from the Stack to be used as th	8		6	• 00	=	• 00		• 00				. 80		000		AMM - B
Source data is located at the a 4.5 post increment R4 or R5, 6—MVO Instruction—post in Other instructions—pre-	SSS, MMM		SSS	MMM, DDD	000	MMM, DDD		MMM, DDD		MMM. DDD		MMM, DDD		MMM, DDD		or of cuclar if &
INDIRECT ADDRESSED DATA—REGISTER WMM Sourcedat all cotate at the address contained in R WMM = 4.5 post increment Re or R5. WMMM = 6—WVO flastraction—post increment R6 PUSK d	MVO@			MVIG		ADD(6		SUB(ii		CMP (s		AND(e	Ī	XORE		MMM 55 real name of management of E-LAA

when the			S. Z. C. 0V				S. Z. C. 0V				S. Z. C. 0V			S. Z				2.2			
IMMEDIATE DOUBLE BYTE DATA—REGISTER Note: The SDBD command is provided by the assembler when the immediate data is greater than the memory width and requires two bytes.	MoVe in immediate double byte data to register DDD. I's will be low hate and It's tener byte XX = dnn't care		ADD formediate double hate date to contents of societer DOD	Results to DDD. L's indicate low byte of literal, U's upper	byte		SUBtract Immediate double byte data from contents of register	DDD. Results to DDD. L's indicate low byte of literal,	U's upper byte		CoMPare Immediate double byte data with contents of register	SSS by subtraction. Results not stored L's indicate low	byte of literal, U's upper byte.	logical AND Immediate double byte data with the contents of	Register DDD. Results to register DDD, L's indicate low	byte of literal, U's upper byte.		exclusive OR Immediate double byte data with the contents	of register DDD, Results to Register DDD, L's indicate low	byte of literal, U's upper byte.	
e: The Squires tv	100	∄	nnn	000	III	nnn	001	000	===	nnn	001	SSS	===	100	000	Ħ	nnn	00	000	Ħ	nnn
ER Not	95	=	nnn	=	Ħ	nnn	000	Ξ	===	nnn	000	Ξ	111	000	Ξ	Ħ	nnn	000	Ξ	∄	nnn
REGISTI	0000	XXEL	OUUU	1011	XXLL	XXND	0000	1100	XXII	NXX	0000	1101	XXIII	0000	1110	XXII	XXDU	0000	Ξ	XXII	XXIN
IMMEDIATE DOUBLE BYTE DATA—REGISTER Note: The SDBD co immediate date is greater than the memory width and requires two bytes.	14		14		J		34				14			14				*			
E DOUBL	1, 000		udu I				1, 000				I. SSS			000 1				000 1			
IMMEDIAT	MVII		Annı				SUBI				CMPI			ANDI				XORI			

MMM—Address Mode 000—Green address in location following instruction 001—Lindings address by Basicas P.	001 — Indirect address for Register 1 010—Indirect address for Register 2 101—Indirect address for Register 4, post increment 100—indirect address for Register 6, post increment 100—indirect address for Register 6, post increment for MVO only 110—indirect address for Register 6, post increment for MVO only indirect address for Register 6, pre-cerement for all instructions except NVO 111—Indirect address for Register 7, post increment. (Immediate data in focation following instruction.)	
GLOSSARY OF TERMS SSS—Source Register DDD—Destination Register non-minimar of stude s	R —Register to Shift lank 0—3 allowed) AAAAAAA Memory address for Jump. AAAAAAAAA (Memory address for Jump. BB—Register to save off of Cin for Jump. (Reg = 188.4, 5 or 6) S —Sign of address displacement for faunt (Por Persive). papping Peperpepper—Address displacement of the manning varieties of the same of the	



COMPONENTS ...

Resistors

R1, R2, R14, R15, R16
R3-R13, R17, R50
R18-R33
R34-R49
R51-R54
R55
R56
R56
R1 k (5 off)
R1k (5 off)
R1k (13 off)
R100 (16 off)
R300 (16 off)
R51-R54
R56
R56
R56

Capacitors

10n ceramic 47p ceramic (2 off) C2. C3 10μ 16V tant. (14 off) C4-C17 470μ 10V elect. C18 220n (2 off) C19, C20 4700µ 10V elect. C21 1000u 10V elect. C22 2200µ 25V elect. C23 100µ 10V elect. C24

Diodes

 D1-D16
 TIL 209 (16 off)

 D17, D18
 IN914 (2 off)

 D19, D20
 OA47 (2 off)

 D21, D22, D23
 IN4001 (3 off)

 D24
 3V3 Zener BZY88

 D25
 6V2 Zener

Transistors

TR1, TR2 2N3705 (2 off)
TR3 2N3904

Integrated Circuits

IC1, IC27 7404 (2 off)
IC2, IC3 7474 (2 off)
IC4, IC7, IC10 7400 (3 off)

7410 (2 off) IC5, IC9 7406 106 74145 108 7402 IC11 IC12 74L30 IC13, IC14 74175 (2 off) 74L75 (4 off) IC19, IC20, IC21, IC22 IC15, IC16, IC17, IC18 2112 (4 off) CP1610 **IC23** IC24, IC25 7805 (2 off) 7403 **IC26**

Switches

\$1 3 pole change over (toggle) \$2, \$3 5.p.c.o. push button (2 off) \$4, \$19 5.p.c.o. centre off (16 off) \$20 5.p.s.t. 1A \$21–\$24 Push button (4 off) \$25 5.p.d.t.

1A slow blow fuse 20mm

Miscellaneous

FS1

T1 240V, 6V 1A, 6V 1A XL1 1MHz crystal (380 x 200 x 100mm) Case Case (150 x 80 x 40mm) Fuse Holder Holders for i.c.s. (if req.) Printed circuit board (main board) Printed circuit board (power supply) LS1 8ohm 0.2W speaker 7 way DIN socket (2 off) 7 way DIN plug (2 off) Veroboard

CONSTRUCTORS NOTE

All the components including the CP1610 microprocessor are available together with a data sheet from **Technomatic** and **Watford Electronics** who both advertise in the magazine.

TS1 when data is not valid. IC8 pin 2 outputs ADAR which causes the addressed contents of memory to be gated on to the bus and strobed into the ADDRESS REGISTER. Pin 5 outputs BAR which strobes the CPU output into the ADDRESS REGISTER. These two low going signals are orred together by IC9a, and used to strobe the address into IC13 and 14, the address latches. IC8 pin 3 outputs IAB which, via gates 11a and b, enables the RAMs IC15 to 18, to output the program starting address on to the bus. IC8 pin 4 outputs DTB. This signal is fed via gates 9b and 11b to the RAMs chip enable to read data out on to the bus. The final output of IC8, pin 7, is DWS. This signal also enables the RAMs via gates 9b and 11b, but it also strobes the RAMs write inputs via gates 9c and 10a. It is also used to strobe data into the output latches IC19 to 22 when gated with address 255 in gate 11c.

ADDRESS AND DATA ENTRY

When S1 is in the load position S1b provides +5V to one bus feeding the switches S4 to S19 while S1c provides 0V to the other bus. Switches S4 to S19 are normally at their centre off position. To load an address into the address latches, first set the switches to the required binary pattern (+5V="1", 0V="0") and then push the address enter button S2. S2 feeds the anti bounce circuit IC7a and b. IC7 pin 8

goes low; this puts a low on IC9 pin 1. IC10 pin 6 goes low, reading the data from the switches into the address latches when it goes high again.

The word that has been loaded into the address latches is now addressing the RAMs although they are not yet enabled. By setting the sixteen switches to the required DATA pattern and pushing the ENTER DATA switch S3, the data is entered into the addressed RAM location. When S3 is pushed IC7 pin 6 goes low and this enables the RAMs via gates 9b and 11b. At the same time it strobes the RAMs write enable via gates 9c and 10a, writing the data from the sixteen switches into the RAMs.

OUTPUT LATCHES

The output latches (IC19–IC22) are 74L75 quad latches. The 16 bus lines feed the D outputs of the latches. When binary address 255 is selected (OCTAL 377) IC12 pin 8 goes low and if DWS goes low at the same time the output of gate 11c goes high, strobing the data from the bus into the latches. The output latches provide both Q (high going) and \overline{Q} (low going) outputs. The \overline{Q} outputs are used to drive single l.e.d.s via current limiting resistors. The Q outputs of the latches could be brought out to drive external circuitry if required.

NEXT MONTH: Construction details

VEHICLE INSTRUMENT DISPLAYS B. Shepherd C. Eng FIEE*

The car market is acknowledged to be one of the most difficult for electronic systems to penetrate. The combined environment of temperature, humidity, vibration, electrical transients, etc. coupled with the required visual appeal and cost targets, are very demanding. Nevertheless, the present state of the available technologies indicate that displays will be available for vehicles in the early 1980s.

THE modern motor vehicle contains instruments that have been optimised to provide good performance at reasonable cost levels. They are usually attractively mounted in plastics mouldings and form an integral part of the vehicle fascia. The content of the instrument package is defined by the vehicle specification, written around the marketing requirements, on the generalised assumption that the cost is directly related to the complexity.

Electronic instruments in the form of tachometers and speedometers, etc. have been introduced in the past few years which utilise medium-scale integrated circuits, but again the display system used is a moving-coil instrument that is essentially of mechanical construction. The manufacturing cost of an individual instrument is fairly low, and to consider utilising electro-optic methods of display on an individual basis would generally be too expensive.

BASIC REQUIREMENTS

Examination of available display technologies indicates that separate instruments, complete with drive electronics, would be more expensive than their conventional electromechanical equivalent. If, however, all the individual elements of a dashboard are combined on one large display, and the electronics are also integrated to form one complete panel, then a cost-effective solution can be realised. The phrase "solid-state displays" has been coined to cover this type of panel, though it can be a misnomer in certain cases.

The most important parameters to be considered when choosing a particular technology are those of overall system cost, visual appeal, reliability, life and legibility, which covers the brightness and contrast aspects. The display must be capable of

being manufactured on a large scale with a reasonable level of tooling and a minimum of assembly labour. Furthermore, the display technology chosen must be compatible with available integrated circuits and operate within reasonable voltage and power limitations. The ability to multiplex becomes a very critical factor.

The main parameters of importance for choosing a display technology suitable for a motor vehicle are those of total system cost, visual appeal, reliability, including operation over a wide temperature, vibration, and humidity range, together with a life requirement equal to that of the vehicle.

DISPLAY TECHNOLOGIES

Generally, the available technologies break down into those suitable for small-area display and those suitable for large areas; they also have different suitability as regards the optimum number of bits. A "bit" of information is defined as a single separate display area that has to be controlled independently (Table 1).

DISPLAY COST COMPARISON

The graphs in Fig. 1 show how the cost of these displays differs depending on their manufacture. A display consisting of only a few bits, e.g. warning lights, will have the lowest cost using individual elements like l.e.d.s. A display with an extremely large number of bits will have the lowest cost using a cathoderay tube.

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TABLE 1. OPTIMUM NUMBER OF BITS FOR A GIVEN DISPLAY TECHNOLOGY

Technology	optimum number of bits
Tungsten filaments Light-emitting diodes	1-20
Cathode-ray tubes	10k-250k
Gas discharge (plasma panels) A.C. or d.c. electroluminescence	30–5k 30–3k
Liquid-crystal display Electrochromic (liquid or solid)	5-200 5-200
Electrophoretic	5–200
Vacuum fluorescent	10–100

A typical full instrument panel for a vehicle will have between 150 bits for a low-line vehicle to 300 bits for a high-line vehicle. It is evident that d.c. electroluminescence (d.c.e.l.) is most cost effective within this range, and that other technologies, e.g. liquid-crystal display (l.c.d.) and vacuum fluorescence, could also prove viable. The system of course also includes the drive electronics, and thus the system cost will involve other technical considerations as described later.

D.C. ELECTROLUMINESCENCE

The particular advantage of d.c.e.l. for vehicle-display applications is that, under normal and low ambient lighting, the display has considerable visible appeal with a reasonably wide range of colours. The disadvantages at present are associated with the available brightness and contrast ratio under extreme sunlight conditions. Work is continuing to overcome these problems.

The technology emits light and hence requires no other means of illumination. The construction is outlined in Fig. 2. The overall visual appeal, together with package cost advantages, have made it the favourite technology at present.

D.C. electroluminescent displays are based on polycrystalline copper and manganese-doped zinc-sulphide power phosphors. The devices are fabricated on transparent conductively coated glass substrates, which are photolithographically etched to provide delineation of the required light emission pattern and electrical input. All displays are hermetically sealed in a robust flat-pack configuration.

As a display technology, d.c.e.l. possesses many unique and advantageous features. The devices operate with high visual efficiency giving excellent legibility and an arresting appearance. The emission spectrum is broad band with its peak close to that of the human eye sensitivity curve. The basic bright yellow colour may be externally filtered to provide readily discriminated green and red displays. The emission is physiologically restful for continuous observation and contrasts sharply with unselected areas of the display and the surround.

Light emission takes place at the junction of the transport conducting film and the phosphor layer. It occurs uniformly over the whole positive electrode area, and individual phosphor particles cannot be resolved with the naked eye.

The display brightness is continuously variable and the voltage range is small because of the superlinear brightness voltage characteristic. There is no threshold or striking voltage. A vast range of display formats is possible. High resolution is provided by photoetching of the substrate conductor pattern. Combinations of complex shapes both small and large may be provided on the same substrate, as may analogue and digital information together with fixed legends.

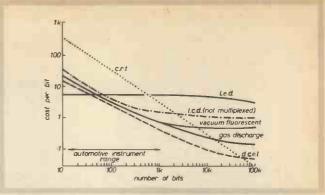


Fig. 1. Comparison of cost per bit against number of bits

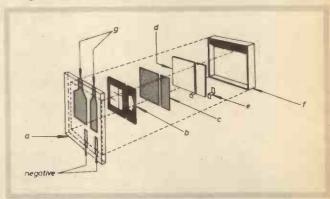


Fig. 2. Assembly for a d.c. electroluminescent display

(a) Front glass.
(b) Insulating mask.
(c) Phosphor, which is in contact with g through apertures in b.
(d) Aluminium rear electrode.
(e) Conductive strip.
(f) Hermetically sealed cover.
(g) Transparent conductive pattern (positive).

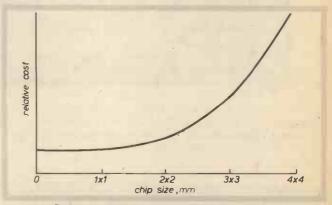
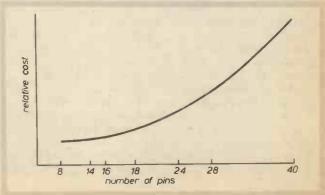
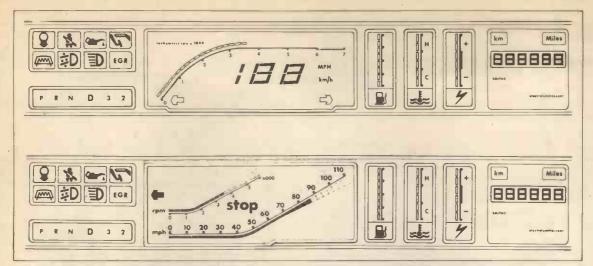


Fig. 3. Integrated circuit cost variation with size
Fig. 4. Integrated circuit cost variation with number of pins





Complete panel styles (a) with digital speedometer; (b) with analogue speedometer

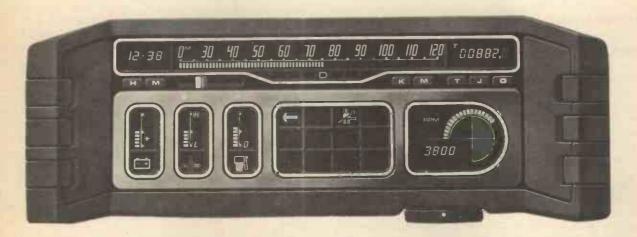
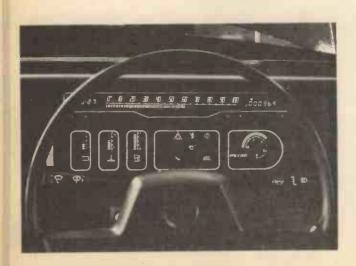


Illustration of a display designed for the Austin Princess, showing (above) all controls and (below) the presentation in the car (see also front cover).



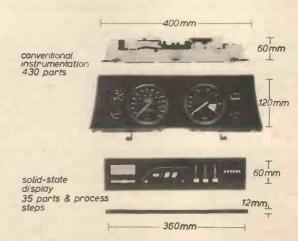


Fig. 5. Size comparison of conventional and solid-state displays

CONSTRUCTION

The simplicity of construction and processing are well suited to large-scale production, and flexibility in design allows economic manufacture of small quantities for prototype assessment. Custom displays are more readily realised than with any other solid-state light-emitting-display technique. D.C. electroluminescent displays are truly solid state. There is no sudden catastrophic failure mode. The slow increase in electrical resistance of the phosphor layer, which is characteristic, may be compensated for by simple constant-power-drive conditions. Neither luminous nor electrical power conversion efficiency show any appreciable decrease during operation.

The target life for this type of vehicle instrumentation is 5,000 hours. Life is measured as the time between the initial brightness level down to one half of this value. Results obtained on small panels are giving encouraging results and continuous life-test data are being collected.

It is expected that such displays will first appear on vehicles with panels of approximately 100cm², containing digital clock, tachometer, fuel and temperature gauges and warning lights. This will be followed by complete dashboards containing speedometers and odometers also.

Simpler instrumentation, such as a digital clock alone, can be more suited to a digital l.e.d., or vacuum fluorescent or gas display panel. These types of displays are now available from many sources, using a standard 7-segment digital format, and could be combined with a conventional instrument to form a clock/tachometer display for instance.

INTEGRATED-CIRCUIT TECHNOLOGY

An instrument control system requires a combination of fairly dense logic for quartz-controlled digital clock, and odometers, linear analogue circuitry for bar-graph displays, e.g. small gauges, tachometers, etc. and display drive circuitry dependent on the type of display required. The complexity of the logic plus the fact that part of the circuit must still operate when the ignition is off will require low power. Electrical interference will require high noise immunity. Many displays require high-voltage switching.

The integrated circuits currently being designed are using integrated injection technology for the logic and high-voltage bipolar circuits for display drive. Custom-designed integrated circuits are considered best for this market because of the optimum product costing relative to the large volume requirement for the complete market.

Partitioning of these circuits has been made both to give the most flexible "kit" from which to build various panels and the lowest cost relative to Figs. 3 and 4.

As the display content grows, it becomes increasingly evident that "multiplexing" is essential to produce the lowest cost and highest reliability. Multiplexing is the term given to controlling the display element by its two connections, in parallel with other elements, so that the minimum number of interconnections are required. A non-multiplexed panel with 256 elements will require 257 connections. Multiplexed in a 16 x 16 manner only 32 connections are required.

Each connection requires a signal from an integrated circuit and a pin connection to that circuit. Typical low-cost circuits have 14 pins with a cost increase for the other standards of 16, 18, 24, 28 and 40. Also, the cost of an integrated circuit is dictated by the area of silicon used, with a cost curve generally proportional to area up to the point where low yield increases the cost still further. Figs. 3 and 4 indicate typical variations in cost.

The set of integrated circuits that will be used is being designed jointly with Plessey microelectronics, and samples are now becoming available for complete system evaluation by various European car manufacturers.

STYLES

The flexibility of a solid-state display to produce an attractive instrument panel is very important. The dashboard is immediately in front of the driver and the display aesthetics can be an important factor not only in the sale of the vehicle, but also in the subsequent safety aspects, and the manner in which information is transmitted from the vehicle to the driver. Early styles are shown on the cover and opposite.

As can be seen from the size information in Fig. 5, there is considerable potential space saving advantage when using solid-state displays.

FUTURE

Further examination of trends indicates that integrated circuits are becoming more complex and larger for a given cost level, and the display technologies are also improving. The cost per bit is therefore falling for large displays, but remaining the same for small displays. From this we can deduce that it will be most economic to offer more and more facilities at the same cost level.

Microprocessors are becoming available that can compute the relationships between several input parameters and give the driver better information. The instrument panel can become the control centre for the vehicle, optimising the vehicle performance for fuel economy, braking, entertainment, etc.

Coupled with information systems, e.g. route guidance, diagnostic sensing of engine and vehicle performance, this control centre will make the driving task much easier, only giving the driver information when he requires it or in an emergency. A keyboard will probably appear, integrating calculator, and a message display panel as an aide memoire for things like shopping lists.

The boredom of a traffic jam may eventually be overcome by electronic games and in-car entertainment could take on a new meaning.

For assistance in all the above display work, acknowledgement is gratefully given to the Royal Signals Research Establishment, Phosphor Products Co., and Plessey I td.

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SCHAID BEN J. DUNGAN.

TO LEE BEN J. DUNGAN.

Part 1—A fully protected two channel sound to light modulator custom designed for mobile work. THIS article describes a sound to light modulator which was developed to overcome the problems found in many units. The design is intended primarily for discotheque applications, therefore emphasis is placed on reliability.

In essence a sound to light modulator converts analogue signals into a quasi-digital format which conveys the overall pattern of the music into modulation of the display lamps. An SLM can give considerable enhancement to the mainly physical class of music danced to in discotheques.

The development of SLMs, it would seem, has been overshadowed by gimmicky effects as found in random lighting displays which add little to the music, as a result they have remained relatively primitive since their inception some ten years ago.

Many discotheque operators will have experienced problems resultant of this poor state-of-the-art. These are summarised below:

- (1) Continual adjustment is needed at every change in programme level in order to maintain the required degree of lamp saturation.
- (2) Poor channel separation, resulting in a confused display.
- (3) Unreliability, for no apparent reason.
- (4) R.f.i. appearing in audio circuitry.

The circuit to be described has been designed according to the premise that simplicity is the first step towards reliability; this being an essential requirement for discotheque equipment. For this reason, the design is not totally devoid of the aforementioned shortcomings, notably the first, but is certainly a step in the right direction.

THE FAULTS DISCUSSED

Before considering the circuit, it is useful to discuss the faults and their solutions. First (1).

A triac has a certain set voltage at which it will trigger. If the device is turned on continuously, the associated lamps will light continuously, resulting in a boring display; this situation is termed "oversaturation". When the signal level is suitably set, the triac will only switch on when most prominent parts of the music (usually the bass line and vocals) are prevalent. In this situation, the overall rhythm of the music is imitated. The associated lamps are then said to be correctly saturated. In common terms, "They flash to the music."

If the signal level is too low, the lamps will be switched on very briefly or not at all. The signal level for perfect saturation is critical, but some ±3dB variation in "mean" voltage will not seriously degrade the display.

There are two ways around the problem. Firstly, the SLM could be provided with a compressor, which will ensure a constant triggering voltage. However, a simple compressor cannot discriminate between long term changes in audio level and short term changes in the dominant sounds. It will tend to bring parts of the mix which were intended to remain in the background to the forefront, and reduce the dominant rhythms to a common level, thereby giving a false display.

A careful choice of time constant can alleviate this to an extent, but complex circuitry is almost certainly involved if a complete solution is sought. Futher, setting up the SLM so that the lighting level remains constant over the anticipated range of sound pressure level requires a precise knowledge of signal levels, again requiring complex circuitry if rapid setting up is necessary. The use of a compressor was not explored because a far simpler solution exists.

If the signal for the SLM is derived prior to any variable level or tone controls, then changes in signal level resulting from the operation of these is no longer a problem. There remains the changes in level resulting from the dynamic

SPECIFICATION

Bass —3dB Turnover Frequency	
	Octave
Treble –3dB Turnover	3.5kHz —18dB/
Frequency	Octave
Input Sensitivity	2V r.m.s.
Input Impedance	>100 kilohms
Input Protection	AC to >60 volts
	(equivalent to
	450 watts into
	8 ohms)
	DC to >90 volts
Channel Power Capacity	1.4kW (220-250
	volts)
RFI Suppression	Zero voltage
	switched,
	choke-
	capacitor
	filtering
Overcurrent Protection	Glass fuses for
	small overloads
	to 80 amps.
	High speed
	fuses for short
	circuits to
and the second second	850 amps
Overvoltage Protection	VDRs and
	snubbers
	against fuse arc
	and mains
	transient
	suppressors

range of the source signal (typically 50dB for albums) and from differing modulation levels (a 12in single having the highest, then 7in singles, and 12in albums with the lowest). However, since we require in any case to capture only the overall pattern of the music, herein defined as the recurrent peaks, the former point can be neglected. Regarding the latter, the 6dB leeway, previously noted, that will permit a satisfactory display is the solution to the problem of differing disc modulation levels.

Careful setting up is required nevertheless; this is speedily effected by testing the display using a noisy 12 in single and a relatively quiet album track.

It must be emphasised, however, that the direct connection of sensitive audio equipment to circuitry employing triacs provides an excellent opportunity for r.f.i. to creep into the audio chain, hence r.f.i. must be reduced to negligible levels.

POOR CHANNEL SEPARATION

Many designs feature 6 or 12dB/octave filters. Further, three channels are often used. Disco music can be adequately presented with two channels, simply bass and treble, and indeed, more channels serve only to confuse.

For two channels to act discretely, yet cover a reasonable range of frequencies, a high order of filter is required; 18dB/octave (3rd order) filters are the compromise between filters of a higher order requiring high tolerance components and lower order filters which would make even a two channel display ambiguous.

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UNRELIABILITY

The most likely reason for unreliability is ignorance of the need for transient overload protection when triacs are used in mains circuits. For all its inherent reliability, the triac is easily destroyed, particularly when operated from the mains supply, where extremely high transient voltages and fault currents can be encountered. The incandescent lamp has two positively unwelcome characteristics:

(a) When power is initially applied to a lamp, a surge current an order of magnitude higher than the nominal current flows, reaching a peak after some 400µS, which lasts for up to 20mS and then dies away exponentially, reaching the nominal current level in 70mS. In an SLM, power is constantly reapplied to the lamps, hence the surge conditions are repetitive.

(b) When a lamp filament ages, it grows brittle and will eventually fracture or "blow". The sudden change of current causes arcing across the fracture, and the heat of the arc may be sufficient to ionise the nitrogen-argon mixture around the filament. The ionised gas will conduct heavily, resulting in a short circuit which lasts until the arc is extinguished. Filament failure in infrequent, but should it occur, an unprotected triac will almost certainly be destroyed. Fuses are the most practical means of protecting triacs from short circuit conditions in mains circuitry. Prior to considering fusing in detail, the most likely overload conditions will be identified; these are:

Small overload. Human error will make it inevitable in the long run that an SLM unit is overloaded. Given that the overload is highly unlikely to exceed 100 per cent, ordinary glass fuses will protect a triac.

Lamp surge. Fourteen 100W lamps at 225V will cause a 77A surge current to flow repetitively in an SLM. Any fuses used, and the triac, must be capable of withstanding this surge on a continuous basis.

Short circuit. The failure of a lamp filament can result in a short circuit and the prospective fault current is limited only by the supply impedance. Assuming worst case conditions, a current of 1,000A peak can be expected on a 30A ring circuit. Since short circuits can also occur if cables are damaged, the relative unlikelihood of filament failure causing a short circuit should not be cited as a reason for disregarding the necessity for short circuit protection.

Because semiconductor junctions can be destroyed extremely rapidly under short circuit conditions, the common glass fuses are inadequate; high speed fuses must be utilised. These have precisely defined characteristics and are capable of clearing high current faults in a matter of milliseconds.

The rapid current cut-off causes arcing and the fuse will not clear the fault until the arc is extinguished, which happens rapidly; the important point to note is that the arc voltage may be considerably greater than the mains voltage.

The fault protection arrangements in many SLMs is nonexistent or pitifully inadequate. This design is protected against the aforementioned overloads.

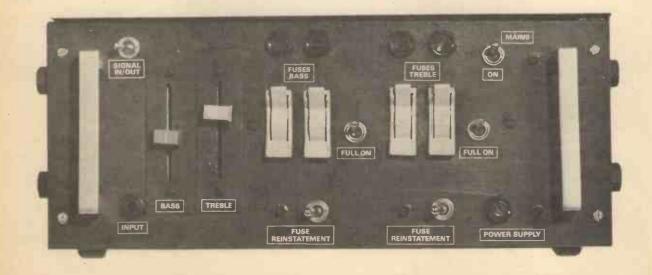
FUSE SELECTION

For adequate discotheque lighting, maximum use of a 13A outlet should be aimed at. Assuming that the minimum mains voltage is 225V, fourteen 100W lamps per channel are permissible with enough capacity to power an amplifier and discotheque console. The lamp surge current using 100W lamps will be 12·4 times the nominal, hence 77A at 225V per channel.

The fuses and triacs must withstand this surge for at least 20mS on a repetitive basis. A special anti-surge type of high speed fuse will be required. A Ferraz 600 CP-AS 10.38/12 fuse will withstand a 78A surge for 110mS. The same fuse has an I²t of 160A²S at 250V and will clear 850A r.m.s. in 2.0mS. A BTX-94 triac has a 190A²S I²t rating for this duration, and can withstand a repetitive 100A surge; hence it is suitable for the application.

Whilst the use of a 25A triac to control a 6A load smacks of gross overdesigning, it can be seen that the choice of triac is determined primarily by the available fuses, the prospective fault current and the lamp surge current.

A slow blow fuse must now be selected to protect from small overloads and to prevent catastrophic destruction of the high speed fuse which can occur when it is subjected to a steady current in excess of its nominal rating (12A in this case). Glass fuses are produced to wide tolerances, thus selection is based upon worst case samples of the fuse, and good samples may be blown. Specifically, the 5A 1½ in fuse detailed in the design will handle the 6·2A nominal load current provided it is not a fast blowing specimen. In fact, the specified fuse will comfortably protect the triac up to 80A.



TRANSIENT VOLTAGE PROTECTION

The triac must also be protected from transient voltages in excess of its rating. A compromise between an ideal yet complex transient suppression circuit (allowing low cost 400V triacs to be used) and a very high voltage triac (needing little suppression but a lot of money) is required.

To protect the triac from the majority of transients, zinc oxide voltage dependent resistors are connected across the devices. These present a high impedance to the usual mains voltage, but present a low impedance to a transient significantly in excess of the normal voltage, thereby absorbing the energy. Fortunately, transient voltages tend to have an inverse amplitude/duration characteristic, which avoids excessive VDR dissipation. Unfortunately, the highest voltage which the VDR will reduce to a safe level is depen-

dent upon the source impedance of the transient, but in this design, transients up to several kilovolts will be clipped to below 600V by the VDR.

Two voltage transients of known amplitude can be identified. The clearing of the high speed fuses creates a 710V arc and the voltage rating of the chosen triac is selected to be in excess of this. The series inductor, used to mop up residual r.f.i., generates back e.m.f.s of the order of 400V. A capacitor in parallel with the inductor acts to define this voltage.

Finally, a CR network (snubber) is placed across the triacs to assist the operation of the VDR by limiting the rate of change of voltage at the leading edge of a transient (particularly that generated by the high speed fuse) giving the VDR time to operate.

COMP	OMPONENTS Potentiometers		rs	
Resistors			10k log	
R1	22k			
R2	1M			
R3	10k			
R4	1k8	Semiconducto	ors	
R5-R7	10k			
R8	22k	IC1-IC5	741C	
R9 R10	470k 22k	TR1	BC184	
R11	1k	TR2-TR3 CSR1-CSR2	BC337	
R12	10k		BTX94—600U	
R13	3k9	D1-D2	BZY88C—12V	
R14	150k	D3-D4 .	1N914	
R15	22k	D5	0A91	
R16	22k	D6	BZY88C—6V2	
R17 R18	470k 1k	D7	OA91	
R19	10k	D8	BZY88C—6V2	
R20	1k	D9-D13	1N5401 BY206	
R21	10k	1C6	+15V, 1A regulator, 7815	
R22	10k	IC7	-15V, 1A regulator, 7915	
R23	47 10W			
R24	47 10W			
R25	Voltage dependent resistor (R.S. 238-457)			
R26 R27	Voltage dependent resistor (R.S. 238-457)	Inductors		
R28	68	L1, L2	500 turns of 19 s.w.g. enamelled copper	
29	Voltage dependent resistor (R.S. 238-457)		wire layer wound on 200 x 10mm ferrite	
R30	68		rod	
All 1W o	arbon film 5% except where otherwise stated	T1, T2	1:1pulse transformer (R.S. 196-369) (Maplin)	
Compaias		Т3	Mains primary: 17-0-17V 1A secondary	
Capacito				
C1 C2	470n polycarbonate 3n3			
C3	68n			
C4	180n			
C5	10n			
C6	10μ			
C7	3n3	Miscellaneous		
C8	3n3	LP1-LP4	Mains neons	
C9 C10	100n 220n	FS1, 3, 5, 7 FS2, 4, 6, 8	Slow fuses 5A 1¼in Fast fuses Ferraz type 600-CP-AS 10.38—	
C11	100n 1,000V	132, 4, 0, 0	12A	
C12	100n 1,000V	FS9	500mA 11in	
C13	220n 1,000V		1 in panel fuseholders	
C14	100n 250V		Ferraz fuseholders type SGR10	
	6 2,200μ 40V elect	\$1, \$4, \$6	S.p.s.t. 15A toggle	
	8 100n	S2	D.p.s.t. 15A toggle	
C19-C2	0 470n 220n 1,000V	S3, S5	D.p.d.t. 15A toggle 19°C/watt heatsinks (2 off)	
C22	100n 1,000V		2.1°C/watt heatsinks (2 off)	
	4 100n		z . c, wateriodistino (o oii)	
	6 250μ 25V elect			

Practical Electronics May 1979

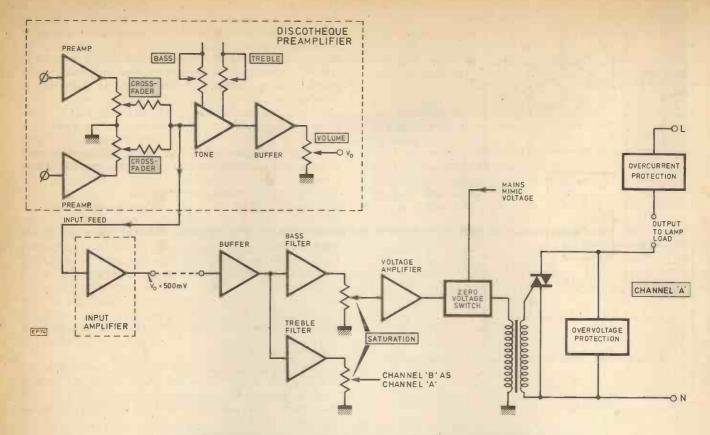


Fig. 1. Block diagram of unit and signal derivation points from typical disco console

R.F.I. APPEARING IN THE AUDIO CIRCUITRY

The use of zero voltage switching is a foregone conclusion when sensitive audio circuitry is directly connected to and in close proximity to a potential source of r.f.i. This is totally eliminated when the triacs are triggered at the zero point of the mains cycle. This is an ideal that is difficult to achieve for two reasons—first the mains voltage is at zero for a very small period of time. Electronic circuitry has a finite operating time. And second, triacs have a minimum holding current below which they will not remain in their on state after removal of the gate trigger signal. For a typical I_H value of 60mA, this implies that the mains voltage for a 1,000W load on a 250V supply must be above 3.7V for reliable triggering to occur. Even at this voltage, not insignificant r.f.i. will be generated. Additional filtering is therefore required.

A series inductor of several hundred microhenries will limit dl/dt to around $1V/\mu S$ (with the nominal load) which is considered satisfactory for reducing r.f.i. on nearby AM broadcast receivers to negligible levels. The addition of a parallel capacitor creates a -12dB/octave low pass filter with a -3dB breakpoint of around 35kHz.

If the unit is housed in a steel, or better still aluminium case, connected to a low impedance earth, and normal precautions are taken with respect to avoiding interaction between conductors carrying small signals and load currents, then to all intents and purposes, r.f.i., whilst undoubtedly occurring, can be disregarded.

INPUT AMPLIFIERS

In order to minimise r.f.i. pickup, the input sensitivity of the SLM is set at around 2V. For reasonably constant lamp saturation, the input signal must be derived prior to volume, tone or other equalisation controls on the discotheque console, mixer, etc. (Fig. 1). Here an input amplifier will normally be required, Figs. 2a, 2b and 2c. If tone or level controls precede equalisation stages, then the input amplifier must be provided with its own equalisation.

In discotheque applications, the microphone may also be required to feed the SLM, and in the case of a stereo desk, left and right channels must be mixed, Fig. 2c.

Power for the op-amps may be derived from power rails in the desk, a separate p.s.u. based on that illustrated in Fig. 4 may be provided, or the input amplifier may be powered from the SLM ± 15 V rails.

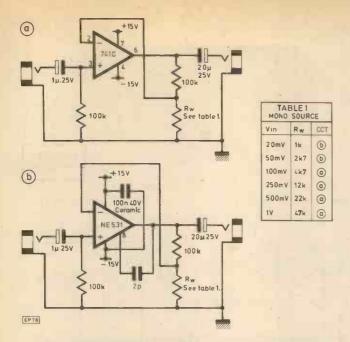
As a last resort, the input amplifier may be mounted inside the SLM, but particular attention must be paid to screening, which should be copper based for maximum effect at high frequencies. The source impedance of stages driving the input amplifier should be less than 20 kilohms.

If the aforementioned signal derivation is too troublesome, a 10 kilohm log pot can be wired across Fig. 3 input and 0V, the wiper connected to C1. The unit will then operate directly from speaker lines at powers of 10 to 450W into 8 ohms and half and double these powers for 16 and 4 ohm systems respectively. The SLM input is ground referenced, so care must be taken to ensure that speaker lines are correctly polarised.

The input is protected up to 60V; below this signal level excessive attenuator settings are not harmful. With this method of signal derivation the lamp saturation is now subject to changes in volume and tone control settings, but its simplicity is expedient and it makes the SLM more versatile.

SIGNAL PROCESSING CIRCUITRY

Referring to Fig. 3, C1 provides d.c. isolation. Zeners D1, D2 and diodes D3, D4 protect the unit from excessive signal input voltages, up to some 60V, in conjunction with C1 and R1.



Figs. 2(a)-(b). Input amplifiers for mono music source

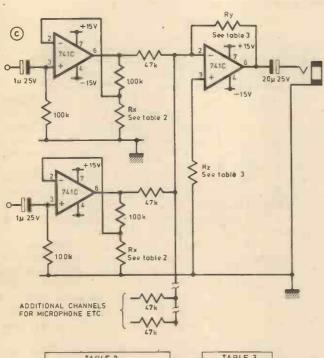


	TABLE 2 MULTISOURCE				
	Vin Rx NOTES				
ı	2mV 1k8				
Ì	10 mV	10 k			
ı	20mV	22 k			
	50mV	47k			
	100mV	DMIT	OPEN CIRCUIT		
	250mV	OMIT	DIVIDE Ry BY 2-5		
	500mV	DMIT	DIVIDE Ry BY 5		

ı	MULTISOURCE		
,	No OF INPUTS	Ry	Rz
1	2	470k	22 k
	3	330k	15 k
	4	220k	12k
	5	180	9k1

Fig. 2(c). Input amplifiers for multisource inputs

Switch S1 is provided to control earthing arrangements. In certain circumstances r.f.i. may be injected into sensitive audio stages, and in most cases the switch will be open to prevent earth loops. If the unit is ever operated from a floating source, however, earthing can be conveniently restored.

IC1 is a buffer-amplifier, providing a low source impedance for the filters and some voltage gain. The filters, built around IC2 and IC4 are bass and treble respectively. These are third order Butterworth types with unity gain in their passband. Their —3dB turnover frequencies are 320Hz and 3.5kHz.

VR1 and VR2 control lamp saturation.

If saturation is seriously upset when changing from 12in singles to albums, or synthesised music with excessive treble content is played, then a switched attenuator at this point can be a useful asset, since it allows rapid correction. The operator must then develop the habit of switching the attenuator when turntable speed is changed.

IC3 and IC5 are voltage amplifiers providing low impedance drive to the zero voltage switch. Zeners D6 and D8 provide a pulsating "strobelike" display, but experimentation with different Zeners or even their omission is suggested, as a lot depends upon the type of display, thermal lag of the lamps and the application.

If one channel is found to be insensitive, then resistor R8 or R15 may be increased. A doubling of value to 47 kilohms is recommended; the corresponding resistor R10 or R16 must then be changed to 10 kilohms in order to minimise offset voltages.

ZERO VOLTAGE SWITCH

A single polarity mimic of the mains voltage is derived from the unsmoothed but rectified output of the power supply. This voltage is applied to the base of TR1 (Fig. 5). When this is above 0.7V, TR1 is saturated and TR2, TR3 remain off. When the base voltage drops below 0.7V (corresponding to a mains voltage of around 11 volts), TR1 turns off and base drive is then applied to TR2 and TR3. The pulse width is around $200\mu S$ symmetrical about the zero point.

The zero crossing pulse is applied at 100Hz to the C106D anode. Two conditions must be met simultaneously for current to flow through the pulse transformer: (a) a triggering signal must be present at the gate of the thyristor, and (b), the mains voltage must be about the zero point.

The C106D thyristor has high gate sensitivity, allowing it to be triggered directly from an op-amp. Whilst pulse transformers may be easily hand wound, the use of a commercially manufactured component which can be relied upon implicitly for reliable isolation is an important safety factor.

MAINS CONTROL CIRCUITRY

Since fuses usually blow at an inconvenient time, a changeover switch is provided in Fig. 4 to allow rapid reinstatement. The blown fuses are isolated by throwing the switch, allowing safe renewal. Neons LP3, LP4 will light when a fuse blows, provided a load is connected. The high speed fuses have a ceramic body, so inspection will not reveal whether they or the glass fuses only have blown. If the glass fuses are intact, then the high speed fuses are "dead" but the reverse is not necessarily so.

The snubber networks C13, R28 and C21, R30 protects the triacs from voltage transients generated by the arcing of the high speed fuses when they clear, together with the voltage dependent resistors in parallel. Protection against mains borne transients is also ensured. C12, L1 and C22, L2

EP79

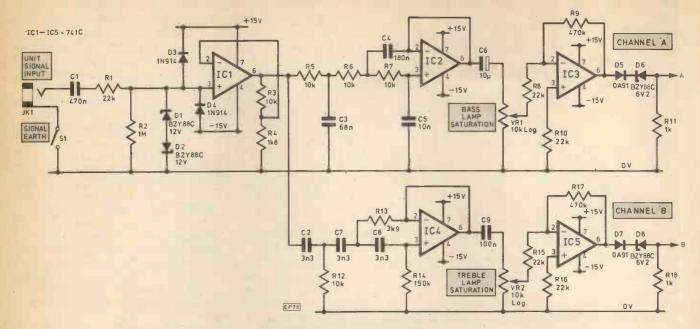


Fig. 3. Signal processing circuitry

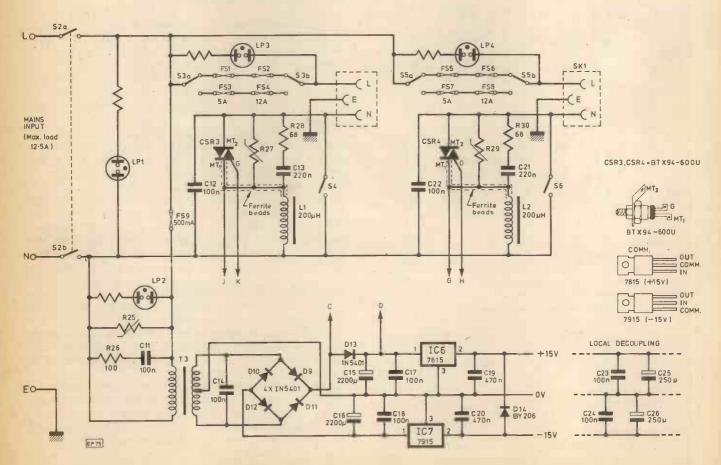


Fig. 4. Control circuitry and power supply

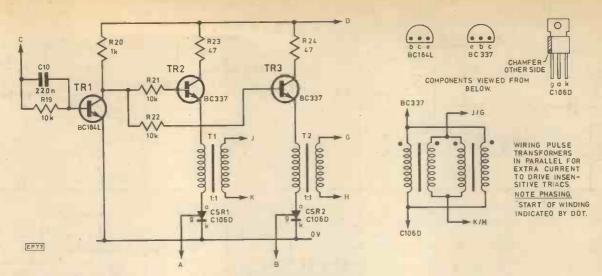
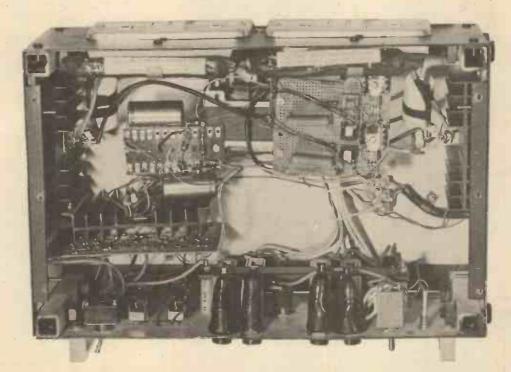


Fig. 5. Zero voltage switch



are placed so that stray capacitances do not hinder the effectiveness of the filtering.

The chokes must be wired as close as practical to the triacs, and the intermediate wire should be covered with ferrite beads (shown as dashed lines).

Mains cables, particularly those carrying the lamp current, should be kept well away from the signal circuitry. The trigger wires to the pulse transformer must be kept short, and should be tightly twisted or in a common sheath. The use of 25A triacs to control a 6A lamp load is undeniably expensive, and if reliability is to be sacrificed for the sake of cost, then 15A plastic triacs (BT 139-600) may be directly substituted. The high speed fuses may then be disregarded.

POWER SUPPLY

R25, R26 and C11 protect the regulators and signal processing circuitry from mains borne and control circuitry

generated transients. Neon LP2 monitors the fuse FS9. The use of a regulated supply is essential as poor mains or transformer regulation may cause interactive effects if the op-amp's supply rails are upset. If both channels switch on simultaneously, a significant drop in mains voltage may be experienced. Also, the protection diodes D3, D4 (Fig. 3) inject supply ripple into the signal line.

Supply ripple is minimised by the use of cheap plastic regulators in place of large capacitors. The BY206 diode further suppresses this effect and allows the regulators to start under a common load.

Local decoupling is provided by 100n capacitors, preferably ceramic types, on account of their low inductance. 250µF capacitors are added to cure possible low frequency instability. Their use is particularly recommended if the supply leads are more than a few inches in length.

Next Month: Construction and setting up.



Transcriptions of a well-known organist's numbers are always interesting. In Harry Stoneham's case they present a definite challenge to brain and fingers as the score is spattered with fast semiquaver and triplet runs. Browsing in my local music store recently, I came across "Play Jazz Organ" containing twelve numbers arranged exactly as Harry Stoneham plays them. These include "Michael's Theme"—his composition used to play in the Parkinson show.

Certain of the numbers in this album have been recorded on l.p. discs in the past and for those that have an organ record collection these are "Hammond My Way" (Contour 2870112) and "Two Fellas to Follow" (Tepee TPRS 100). The transcriptions tally even to the key, which at least simplifies the task of learning to play the score. The album is distributed exclusively by Music Sales Ltd., 78 Newman Street, London W1P 3LA and costs £2.50.

There is a tendency to think that all organs sound the same when fast Leslie is applied but there is a decided difference between the two records mentioned, the first being on Hammond and the second on Lowrey with Automatic Orchestra Control. Harry's fingers are nimble enough but A.O.C. gives an added interest to his brilliant runs.

B-3000

New models constantly appear but the Hammond B-3000 first shown at the NAMM Exhibition in Chicago late last year deserves a special mention. Readers with greying temples will look at our illustration



and think they recognise the Hammond B-3, which was in its heyday some two or three decades ago but is now out of production though still in demand. The fabulous Ethel Smith, some may remember, featured the B-3 in playing "Tico Tico" in a film years ago: she was present at the NAMM show and no doubt the sight of Hammond's latest model brought back memories for her.

In fact, the B-3000 is the electronic equivalent of the B-3 tonewheel organ. Even the consoles are identical at first glance, but there are important differences other than the method of tone generation. Drawbars, second and third harmonic percussion and reverse-colour preset keys are as before but Grand Piano, Electric Piano and Drawbar Sustain have been added. The new organ also has a Transposer which can shift the compass up two semitones or down four.

The purist may say that this device is not necessary, but there is no doubt that playing in D_b sounds brighter than the key of C: the Transposer saves the mental problems of sorting out accidentals in a different key. "Bright Wave" drawbars are available on the lower manual "B" preset, allowing the player to select five flute pltches and/or four bright voices.

IT CLICKED

Certain new rocker tabs have been added, the most interesting being one marked "Key Clicks". This ought to bring a wry smile to the faces of constructors who have burnt gallons of midnight oil trying to eliminate key clicks—but such is the state of the art!

The Hammond's original popularity was partly due to the fact that rhythmic performers found key clicks to their liking (though erstwhile tone cabinets were designed to suppress clicks as far as possible). As the particular tab was fitted during the later stages of design, this special Hammond feature appears to be as popular as ever.

Donald Leslie heard the organ when it was on the stocks and decided that a special tone cabinet was required. The Leslie 822 was thus designed to mate with the B-3000 and provides 215W output power—sufficient for a large audience and making the installation ideal for the professional performer.

One disadvantage of the original B-3 (and indeed the C-3, A-100 and RT-3

organs) was that the preset keys required only a light touch. It was only too easy to take off the "B" preset when "smearing" the manual, so being left with no drawbars in operation—an embarrassing episode in public performance!

The new organ has a more definite preset action and it is now possible to combine several keys simultaneously for multiple combinations.

Priced at around £5000, the B-3000 does not have a rhythm unit, automatic chords or arpeggiator, but two full 61-note manuals and a 25-note pedal clavier make it suitable for playing classical and light music. It makes a pleasant change to be able to report that this is a real organ with the minimum of frills, ideally suited to the professional musician.

TAPE-FOR GOOD MEASURE

Many constructors never manage to learn to play the instruments they make and conversely others that can play go aground with electronic problems. Even so, there are those who are fortunately accomplished in both fields.

Whatever the level of musical ability, the best critic is oneself—heard on a tape playback. This method of self-discipline is excellent for—as the reader has probably found—the simple act of pressing the "Record" button will be followed by a number of disasters! The player Is under pressure, as he would be if suddenly confronted with a public performance, but at least the tape can be scrapped: better still, the tape is used to spur the player to eradicate the problems.

These points came to the fore recently when talking to a keen young musician who agreed, but pointed out that purchasing his instrument had not left much spare cash for luxuries such as good stereo recording equipment. Having ascertained that he had a stereo amplifier and might be prepared to spend some £30, I decided to work on the project. The following notes may be of interest to anyone in a similar position.

INSERTS

The problems with home recording are not only associated with musical ability: there is also the question of clear, undistorted recording.

Our young friend lives close to both TV transmitters in London, so r.f. breakthrough had to be considered. Consquently, a pair of Grundig Electret microphone inserts were purchased for some £3. They were found to give good frequency response and did not invite a background of TV sound while in operation.

The Electret inserts have an integral FET preamplifier which requires a positive supply of between 3V and 6V, easily found from mercury cells as the current drain is small. The insert, slide switch and cells were mounted in a 6in. length of metal tube, the completed microphones being fitted with a two yard coaxial cable and miniature jackplug. Results from these microphones were far superior to the usual "dynamic" cassette types.

CASSETTE

Both inserts and cassette mechanisms appear in the advertisements in this magazine. Having studied the specifications, a mechanism with mated electronics, twin VU meters, high frequency erase, auto stop and CRO, switch was purchased.

Having too many pending jobs already, I was pleased that the work entailed (other than a cabinet) took very little time. This involved fitting a mains transformer and two other components and arranging input and

output connectors for the stereo amplifier. Suitable series resistors for the high level input were found to be 100k.

The overlay panel supplied was somewhat larger than necessary and was sawn down slightly and fitted with a mains switch and pilot before it was attached to the mechanism and boxed. There was one problem with some brass studs moulded in to the underside of the top panel which were 3BA, so that 4BA nuts had to be tapped accordingly.

The result was a spendid little machine, fully equal to expensive counterparts. After experimenting with various microphone positions, we obtained extremely clean and accurate recordings. Given good tape—Superferric or CRO2—there was no noticeable hiss from the i.c. circuitry, which is non-Dolby, of course. The combination of high-quality Japanese mechanism and Electret inserts should satisfy the limited budget reader. Does anyone want a quarter-track reel to real recorder cheap?

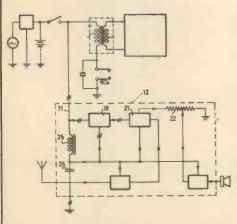


PATENTS REVIEW...

Copies of Patents can be obtained from: the Patent Office Sales, St. Mary Cray, Orpington, Kent

Price 95p each

CAR RADIO A.G.C.



Philips in BP 1 522 421 patent an interesting new approach to automatic volume control for a car radio. The patent was applied for in 1975 and is granted under the old laws. As any driver knows, when vehicle speed increases, so ambient noise also increases. It thus becomes necessary to increase the radio volume to maintain audibility and programme enjoyment.

Many types of automatic speedcompensated volume controls have been proposed in which a gain control signal is derived from a sensor which monitors the engine speed. Such systems require modification and sophistication of the car electric system.

Philips aim to simplify installation by making the audio amplifier gain dependent on the output of an electronic detector which senses ignition pulse frequency. In car radio 12 a frequency detector 18 is coupled to the car d.c. power supply and to an audio control amplifier 21. The supply line carries d.c. and ignition interference pulses; capacitor 26 prevents the d.c. reaching chassis and inductance coil 25 presents a large impedance to the pulses so that they pass to the frequency detector 18.

The detector has a threshold which is continuously readjusted to the value of the last pulse having the largest amplitude and produces a d.c. voltage of which the magnitude represents the frequency of the pulses. This d.c. voltage is fed to the control input of amplifier 21.

The car driver manually sets the car radio volume to a preferred level by means of a conventional control pot 22. Thereafter the signal derived at 18 from the ignition pulse rate, and thus from the engine r.p.m., controls the amplification factor at 21 to vary the audio signal level above and below the mean value set at pot 22.

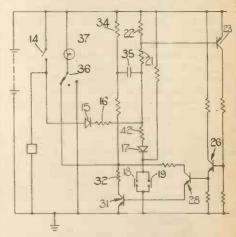
In this way an increase in engine speed and ambient noise produces a compensatory increase in audio level and a drop in speed produces an audio reduction. This produces the subjective effect of constant audio level. The patent gives details of the frequency detector circuit, complete with suggested values for all components.

DELAYED COURTESY LIGHT

Lucas Electrical in BP 1 513 127 (granted on a 1974 application under the old laws) patent a simple modification of a car electrical system which keeps the courtesy light illuminated for a predetermined period after the door is closed—but only until the engine is started. Normally, of course, the lamp is switched directly on and off by opening and closing of the door.

When a door is opened, one of two switches 18, 19 closes and capacitor 35

charges through resistors 34, 42, diode 17 and the closed switch. Transistor 23 is turned on by current flow through resistors 22, 21, 42 and diode 17. Transistors 26, 28 and 31 also turn on and the courtesy lamp 37 is illuminated via a manual switch 36, resistor 32 and transistor 31.



When the car doors are closed, so that switches 18 and 19 are open, capacitor 35 discharges through resistor 21 and the base-emitter of transistor 23 to hold the transistor 23 on for a limited period of time, e.g. eight seconds. During this period transistors 26, 28, 31 are held on and the courtesy lamp remains lit.

If, after the doors have been closed and the switches 18, 19 opened, the car ignition switch 14 is closed (to start the cat) the capacitor 35 discharges rapidly through diode 15 and resistor 16. The courtesy lamp 37 thus goes out almost Immediately.

In this way there is no risk of the courtesy lamp remaining lit while the car is driven off.

Semiconductor UPDATE....

FEATURING: TDA 1028, TDA 1029 MC 3423

R.W. Coles

A.F. SWITCH

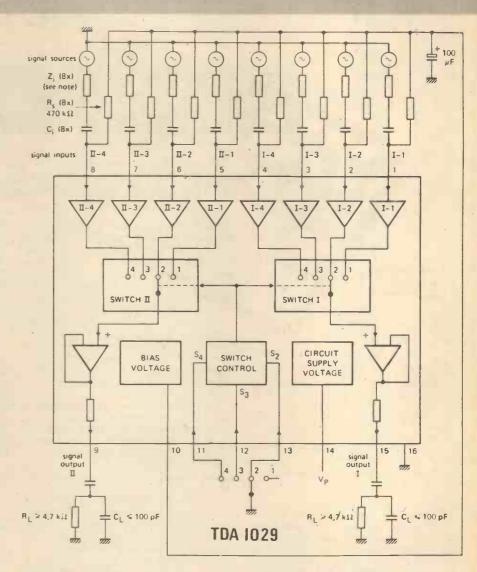
Ever had the problem of switching one of several audio signals to a common output? Traditionally, you would use a few feet of screened cable and a panel mounted switch to do the job. Of course, you would have to use a high quality switch and be very careful with your cable routing, but if the extended signal paths didn't increase crosstalk and mains hum pick-up too much, the mechanical switch solution might be quite acceptable.

More recently electronic switching has become possible. Using this method a mechanical switch is still needed, but now it carries not the signal, but what are in effect logic levels to control the electronic "business end". You could throw one of these together with a handful of op amps and an analogue multiplexer, or you may be able to lay your hands on some op amps whose outputs can be gated into a high impedance state so that the multiplexer is not needed. In either case the solution may be rather expensive and will involve several packages.

To solve the problems of both mechanical switches and the multi-package electronic switch, Mullard have introduced a couple of devices which they call "Signal Sources Switches". The TDA 1028 and the TDA 1029 as they have been coded, look ideal for use in audio, even hi-fidelity audio, systems. They contain, in single 16 pin d.i.l. packages, complete electronic switching systems which run from single 6-23V supplies and feature input protection by clamp diodes together with full output short circuit protection. The TDA 1028 consists of two, two-pole, two-way switches, which means that it has eight input lines and four output lines which are controlled in pairs by two "Switch Control" inputs. The TDA 1029 consists of one, two pole, four-way switch, so it has eight inputs and two "Switch outputs controlled by three Control" input lines. (A fourth control line is not needed because with the three control lines floating, channel one is automatically selected.)

The typical specifications for these devices look good. Over the 20Hz to 20kHz band, the r.m.s. output noise voltage is only 5µV, and a 5V input signal in the same band suffers only 0.04 per cent total distortion. Amplitude response across the band is within 0.1dB, and at 1kHz crosstalk* between inputs is 75dB down.

Apart from use as stereo quad switches, these devices would seem ideal for use in disco systems and synthesisers etc.



CROWBAR

Ever had the experience of a voltage spike on the supply rail which ruins your delicate i.c.s? What you need is a *crowbar*. (No, not to smash up the errant power supply you idiot!)

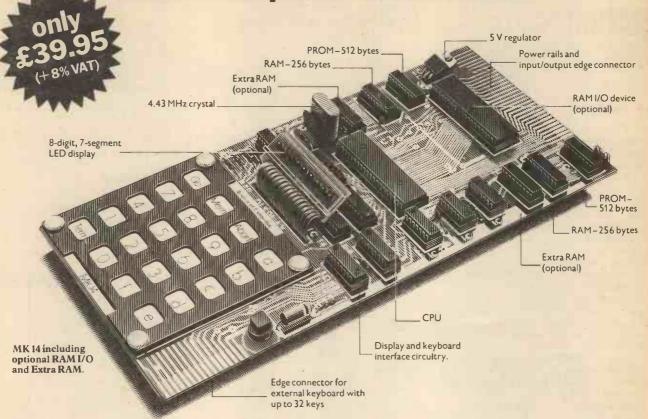
A crowbar is usually formed by using an SCR connected between the raw d.c. supply and ground. When an overvoltage condition is sensed, the crowbar SCR is triggered and the resulting short circuit causes an immediate drop in the voltage supply, followed by a blown fuse within a few tens of milliseconds for a more permanent solution!

The SCR part of a crowbar is simple. The difficult bit is sensing an overvoltage and generating the necessary trigger pulse, two jobs done very well by a new device from Texas Instruments, coded MC 3423.

The MC 3423 lives in an 8 pin mini-chip package and contains a voltage reference and a comparator, together with output drive transistors for (a) the SCR gate and (b) an indicator lamp or l.e.d. which must be fed from a separate supply.

The indicator is necessary because it shows the unlucky user just what has gone wrong. Without it, he would just have to suck his thumb while the wisp of smoke from the fuse drifted skywards!

From Science of Cambridge: the new MK 14. Simplest, most advanced, most flexible microcomputer – in kit form.



The MK 14 is a complete microcomputer with a keyboard, a display, 8 x 512-byte preprogrammed PROMs, and a 256-byte RAM programmable through the keyboard.

As such the MK14 can handle dozens of user-written programs through the hexadecimal keyboard.

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- PROM programmer and blank PROMs to set up your own pre-programmed dedicated applications.

All are available now to owners of MK 14.

A valuable tool - and a training aid

As a computer, it handles operations of all types—from complex games to digital alarm clock functioning, from basic maths to a pulse delay chain. Programs are in the Manual, together with instructions for creating your own genuinely valuable programs. And, of course, it's a superb education and training aid—providing an ideal introduction to computer technology.

SPECIFICATIONS

- ●Hexadecimal keyboard 8-digit, 7-segment LED display ● 8 x 512 PROM, containing monitor program and interface instructions
- •256 bytes of RAM 4 MHz crystal 5 V regulator Single 8 V power supply Space available for extra 256-byte RAM and 16 port I/O Edge connector access to all data lines and I/O ports

Free Manual

Every MK 14 kit includes a Manual which deals with procedures from soldering techniques to interfacing with complex external equipment. It includes 20 sample programs including math routines (square root, etc), digital alarm clock, single-step, music box, mastermind and moon landing games, self-replication, general purpose sequencing, etc.

Designed for fast, easy assembly

The MK 14 can be assembled by anyone with a fine-tip soldering iron and a few hours' spare time, using the illustrated step-by-step instructions provided.

How to get your MK 14

Getting your MK 14 kit is easy. Just fill in the coupon below, and post it to us today, with a cheque or PO made payable to Science of Cambridge. And, of course, it comes to you with a comprehensive guarantee. If for any reason, you're not completely satisfied with your MK 14, return it to us within 14 days for a full cash refund.

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I enclose cheque/money order/PO for £	(indicate total amount.)
Name	0
Address (please print) PE 5 Allow 21 days for delivery.	Science of Cambridge

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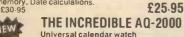
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Approx. 4000 hours continuous use from 2 AA size batteries. Trigs, logs, exponentiations etc. Statistical – stan-dard deviations. Polar to rectangular and R-P. Sexagesimal to decimal Conversion. ENGineering key, Pi, cube root, 6 levels of parenthesis Full memory, \$ x 3 x 5% inches RRP £21-95





£16.95



CASIO MQ-11

Microcomputer Quartz

Specification as AQ-2000 but with "kiss" touch keys and smaller case, $\frac{1}{18} \times 2\frac{1}{8} \times 3\frac{1}{9}$ ins. Leatherette wallet with window for display. RRP £34-95

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 Optional 12 hour with am/pm or 24 hour clock display.
- 24 hour alarm setting.
 Optional hourly chimes
 Chrono measures from
- 1/10 second to 6 hours.

 Net, lap. 1st and 2nd place.

 7.8 mm stainless steel case.
- Mineral glass face

Water resistant to 100 feet.
Less than + 10 seconds/month. Only £39.95



50QS-17B REAL QUALITY AT LOWEST EVER PRICES



Stopwatch, Dual Time. Water Resistant to 66 ft. Displays hours, minutes ten seconds, seconds flash, am/pm; And with day, date and month calendar. Stopwatch measures net times from 1 sec. to 13 hrs. Stainless steel encased (8mm) Fine stainless steel bracelet with easily removable links Mineral glass face.

DOWN

SPORTS WATCHES



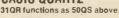
52QS-14B Right, 8mm £22.95



Up to 25 functions. Net, lap and first and second place times to $\frac{1}{100}$ sec. F-100 Resin case and strap. 52QS-14B Stainless Steel encased version with fully adjustable s/s bracelet



CASIO QUARTZ







31QR-20B Left, 8-5mm £19.95

54QS-16B Right, 7-5mm 6 digits £24.95

6 DIGIT WATCHES (Not Sports). Hours, minutes seconds and day of week. (Model 54QS has an optional display of hrs, mins, date, day, ten seconds, seconds flash); And day, date, month, year calendar. Selectable 12 hr (with am/pm) or 24 hr clock.

DOWN

CHRONO AND ALARM









CHRONOGRAPH. 6 digits as above, with chrono measuring net, lap and 1st and 2nd place times from the second to 6 hours. Dual time facility. ALARM. Displays hours, minutes, seconds (or date) day, am/pm. Perpetual calendar. 24 hour alarm.

Cambridge CB1 1DB

WORLD RADIO TV HANDBOOK 1979

Price: £9.16

Price: £6.35

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HOW TO BUILD A COMPUTER CONTROLLED **ROBOT** by T. Loofbourrow Price: £5.20

HOW TO REPAIR VIDEO GAMES by R. Goodman

TESTING METHODS & RELIABILITY **ELECTRONICS** by A. Simpson Price: £4.50

CLOSED CIRCUIT T.V. FOR TECHNICIANS VOL I by K. J. Bohlman Price: £5.00

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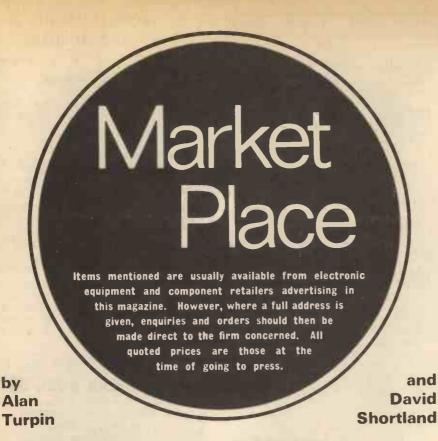
Solve your communication problems with this in robust plastic cabinets for desk or wall mounting. Califally falk/listen from Master to Subs and Subs to Master. Ideally suitable for Business, Surgery, Schools, Hospitals and Office. Operates on one 9V battery. On/off switch. Volume control. Complete with 3 connecting wires each 66ft. A Battery and other accessories. P. & P. 399.

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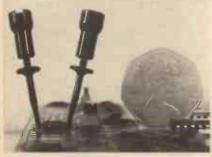
Latest transstorised Telephone Amplifier with detached plug-in speaker. Placing the receiver on to the cradic activates a switch for immediate two-way conversation without holding the handset. Many people can listen at a time. Increase efficiency in office, shop, workshop. Perfect for "conference" calls: leaves the user's hands free to make notes, consult files. No long waiting, saves time with long-distance calls. On/off switch, volume control, conversation recording model at £19-95 + VAT £1-60. P. & P. 90p. 10-day price refund guarantee on all items.

WEST LONDON DIRECT SUPPLIES (PE 5) 169 KENSINGTON HIGH STREET, LONDON, W8



GETTING HOOKED

This micro hook is ideal for test hook ups to i.c.s. Weighing less than one gram it will not damage an i.c. leg while remaining tenaciously attached.



Overall length of the hook is 1.75in and it operates with a "hypodermic" action.

Available ex-stock in ten colours the price is 62p plus VAT and p&p. Cash with order, otherwise minimum order charge £10. Trade enquiries invited.

Special Products Division, British Central Electrical Co. Ltd., Briticent House, New Street, Ringwood, Hants. (04254 4611).

FOR MAINLINE READ ARROW

Arrow Electronics, makers of the "Leader" range of kits, have acquired the entire stock of Mainline Electronics, previously suppliers of amplifiers, audio equipment spares, semi-conductors and components.

It is thought that production of the Mainline amplifiers will be continued and merged with Arrow's existing range of "Leader" kits.

Arrow Electronics Ltd., Leader House, Coptfold Road, Brentwood, Essex CM14 4BN (0277 219435).

PRETTY POLY

A new live performance synthesiser employing patch switching has just been introduced by EMS. Named the Polysynthi this colourful instrument has a four octave keyboard but can cover nine octaves in six overlapping ranges. Tuning is claimed as completely stable.

Outputs from the instrument are polyphonic, pressure dependent control and position dependent control. The first control relates to pressure applied to the keyboard and the second to the highest note played. Their voltages can be routed to external devices. In the photograph the left two panels control the envelope follower, two voltage controlled i.f. oscillators and two ADSR envelope



generators. The centre panel provides the sound source; pulse, square and triangle; noise generator and external input. These can be mixed in any proportion. The two right-hand sections contain the treatment with two and four pole filters and a v.c.a. Included is an analogue delay line with voltage control of delay and with variable feedback and mix. The analogue delay line can be used for echo, reverberation, chorus and flanging. The instrument is expected to retail for around £1,000.

Electronic Music Studios (London) Ltd., The Priory, Great Milton, Oxford.

DIGITAL BREADBOARD

The Powerace 102 is a complete circuit building system that allows prototyping and experimentation without the use of a soldering iron. With a regulated 5V power supply built in, the unit has been designed specifically for digital work. Two breadboard strips allow i.c.s and/or discrete components to be plugged in directly. Both strips have a matrix of 256 terminals each with five interconnected contacts.

The distribution system comprises 16 buses each consisting of 25 electrically connected contacts. These buses may be patched around



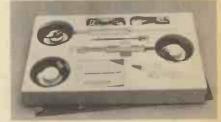
or jumpered together in any combination to provide unique or common functions, such as voltage and ground lines, reset lines, clock lines and shift command.

Other built-in features include pulse detection (100ns high or low selection); two logic switches with debounce circuitry; four data switches (logic 1 or 0); logic indicators (3 buffered l.e.d.s); clock generator offering 1Hz, 10Hz, 10Hz, 1kHz, 10kHz and 100kHz O/Ps; and/or one-shot (7ms, Q or Q)—both Q and Q can source 400µA and sink 15mA.

The price of the Powerace 102 is £105 inc. VAT and p&p. For further details contact Lektrokit Limited, Sutton Industrial Park, London Road, Earley, Reading, Berks RG6 1AZ (0734 669116).

THERMOCOUPLE PROBE KIT

This kit by Comark comprises—a surface probe for medium to large surfaces, with a spring loaded tip enclosed in a ceramic sheath to minimise air current errors, measuring with good repeatability to +750°C; two fast response bead thermocouples with an operating range from -100°C to +250°C; a needle probe handling from -100°C to +600°C; a thermocouple adaptor lead; a two metre extension lead; a quantity of thermally conductive compound packed in a syringe; and a supply of self-adhesive pads.



Price inc. of VAT and carriage is £53.46.

Comark Electronics Ltd., Brookside
Avenue, Rustington, Sussex BN16 3LF
(09062 71911).



Two hundred and six stands, a warm dry atmosphere, and beer at 90p a pint, that was the prospect on entering the Grosvenor House Hotel, Park Lane, on the 27th of February. The exhibition was sponsored by New Electronics and well organised by the Evan Steadman Communications Group.

SMART CASES RULE—OK

Proven in use in the U.S. of A. these project cases by OK Machine and Tool are what any self respecting project builder should be treating himself (herself) to as soon as a circuit idea has been breadboarded and a layout size

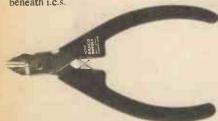


is known. They are available in several widths, heights, and depths in beige, blue or black. The 3mm ABS mouldings incorporate vertical circuit board guides in 11 places, and handles and tilt leg stands are options.

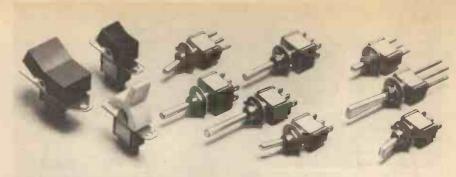
Pac Tec "C" series, from £5.33 by OK Machine & Tool (UK) Ltd., 48a The Avenue, Southampton, Hants SO1 2SY (0703 38966). Trade enquiries invited.

NO SHORTS

With a smart case to house your project what better next than to lash out on a completely new pair of pliers with an off-cut retaining device specially designed to prevent the inevitable short in the hidden areas beneath i.c.s.



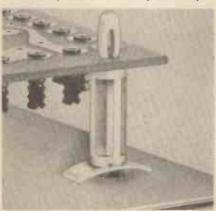
The Swedish company Bahco make a useful range of high quality cutters and pliers, and the 2131 electronics pliers shown are the result of a detailed international survey of industrial users requirements. They have a new design of return spring and the handles are nice and wide. The jaws are available single or double bevelled. These cutters retail at £6.38 plus VAT and Bahco tools are widely available.



TOGGLES, ROCKERS 'N' LEVERS

Pye Electro-Devices have introduced a range of miniature switches; toggle, rocker and lever switches with conventional terminals, wire-wrap or p.c.b. mountable terminals; p.c.b. mountable toggle switches with terminals bent at 90°; p.c.b. mountable toggle, rocker and lever switches with support brackets. Pye are also making an offer you cannot refuse—a sample pack of five different switches at a special introductory price of £2 post free. With a number of types giving thousands of combinations they are obviously not able to select five switches of your choice but the pack has been made up of the most commonly used types and includes one rocker, one lever, and one p.c.b. mountable with edge actuation. This is a limited offer so check availability by phone first. No order required.

Pye Electro-Devices Ltd., Exning Road, Newmarket, Suffolk CB8 0AX (0638 5161).



SUPPORT YOUR P.C.B.s

Are you still fiddling about with nuts, bolts and spacers to support your p.c.b.s and running into the snag of having to avoid any track being close to mounting metal? Nylon circuit board supports are the answer—quick to fix, boards easily lifted off for fault finding, no chance of shorts.

From Chicago, the Richco Plastic Company manufactures fasteners for industry and their stock includes a range of p.c.b. supports; spacers; circuit board guides; wire saddles; cable and wiring accessories; round and flat cable clamps; clips; bushings; extrusions; nylon rivets; circuit board ejectors and pullers.

Mainly a supplier to industry, with a £10 min. order, you may be able to get samples. Stand off distances are from 5 to 22mm. Or persuade your component supplier to stock.

Richco, Universal House, 10a Gilbert Road, Belvedere, Kent DA17 5DA (032 24 47227). Trade enquiries invited.

HOLD 'EM DOWN

One of the most common poorly secured parts inside project boxes which arrive at the P.E. offices is the battery, often wedged in one corner with a piece of cardboard, taped to the side of the box, or even just left to rattle around. The difficulty is to hold the thing firm while still allowing it to be replaced easily. Double sided sticky pads are an ingenious quick fix solution but for bother free replacement times a properly designed mechanical device gives a professional finish.



The holders shown secure those most difficult of d.c. sources, the single cylindrical cell. If more than one cell is needed the nylon holders are easily interlocked to form an array.

Cambion Electronic Products Ltd., Castleton, Sheffield S30 2WR (0433 20831). Trade enquiries invited.

WHEN TWINS ARE WELCOME

The Weir range of Twinpack power supplies contains three models: 0-30V @ 1A, 0-30V @ 2A and 0-60V @ 1A. A current limiting circuit enables the user to set the



current limit to any value from the rated minimum down to zero. Prices, £150-£180.

Weir Instrumentation Ltd., Durban Road, Bognor Regis, Sussex PO22 9RW (02433 26611).

Having built your project go to the next page to find a suitable DMM for testing and fault finding, if you're not already constructing our Autoranging DMM.

The All Electronics Show had a wide range of digital multimeters on display with Gould Electronics launching their Advance Alpha IV, a 3½ digit unit using liquid crystal display. This is their first Alpha type multimeter to feature a l.c.d. and it is claimed that the battery life is increased from 8 hrs. (using the l.e.d. display) up to 400 hrs. The Alpha IV has a basic accuracy of 0.2 per cent. The large liquid crystal display gives automatic indication of decimal point position, polarity and over-range. The instrument normally operates from four "C" cells, but a mains battery eliminator option is also available.



A total of 25 measurement ranges are provided, covering d.c. voltage measurements from 200mV to 1kV, a.c. voltages from 200mV to 750V r.m.s., d.c. and a.c. currents from 200μA to 2A, and resistances from 200μ to 2MΩ. Maximum resolution is 100μV, 100nA and 100mΩ. The unit is supplied complete with test leads, a handbook and a set of batteries. Optional accessories include the battery eliminator unit, a high-voltage probe and a carrying case. The price of the Alpha IV is £105 plus VAT, from Gould Instruments Division, Roebuck Road, Hainault, Essex 1G6 3UE (01-500 1000),

BACH-SIMPSON

The Bach-Simpson 463 has a 3½ digit 0.5in. l.c.d. display with push button selection of all ranges and functions.

The instrument has 26 ranges, 5 d.c. up to 1kV, 5 a.c. up to 600V, 6 resistance up to $20M\Omega$, 5 a.c. and d.c. current ranges up to 2A, with an accuracy of 0-2 per cent on d.c. voltage ranges.



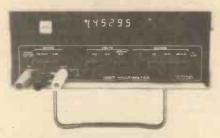
The 463 is supplied with alkaline batteries which enable 200hrs of continuous operation. An automatic indicator gives a warning of low battery condition. The cost of the 463 is £125 plus VAT and carriage.

Bach-Simpson (UK) Limited, Trenant Industrial Estate, Wadebridge, Cornwall PL27 6HD (020881 2031).

51 DIGIT

The Datron 1057 is a high quality multimeter with a $5\frac{1}{2}$ digit display offering a scale length of 199,999. This enables d.c. measurements to be made to an accuracy of 0.01 per cent.

The instrument has 5 d.c. voltage ranges (0-1,000V), 4 a.c. voltage ranges (0-700V) and 5 resistance ranges $(0-20\text{M}\Omega)$. A 1kV peak can be applied continuously to any a.c. or d.c. voltage range and up to 250V r.m.s. on



any resistance range without damage.

The price of the 1057 is £650 plus VAT and a current shunt set is available as an optional extra for £150.

Datron Electronics Limited, Hurricane Way, Norwich Airport, Norwich NR6 6JB (0603 404824).

LAWTRONICS

The Lawtronics range of LM Multimeters feature four l.e.d. models (3, $3\frac{1}{2}$ and two 4 digit) and two l.c.d. (3 and $3\frac{1}{2}$ digit) models. All six models have 21 ranges and are housed in the same size case (48 × 68 × 100mm) making them ideal for field work.



The l.e.d. models are supplied with rechargeable cells and a mains adaptor whereas with the l.c.d. models they are an optional extra. Other extras include a high voltage probe (45kV), leather carrying case and a panel mounted flange case to replace the tilt stand. The prices of the LM range vary from £81 to £195 plus VAT and p&p.

Lawtronics Limited, 139 High Street, Edenbridge, Kent TN8 5AX (0732 865191).

ANALOGUE/DIGITAL?

If you have ever unsuccessfully tried to decide between an analogue or digital system then the Normameter D could be your ideal choice of multimeter. This instrument combines the advantages of both the digital and analogue display in one unit.

The 3½ digit display allows exact readings

to be taken whilst the analogue display shows the changing state of quantities being measured. The digital unit has a maximum settling time for d.c. quantities of 2 secs. A storage circuit is used for smoothness of operation with digits changing only after an integration period. The display car be fixed



and released at anytime by operation of a storage key without affecting the analogue display.

At the back of the instrument there is a 24 pole socket for connection to a printer. The cost of this schizoid multimeter (£462 plus VAT) might well help you make up your mind.

Cropico Ltd., Hampton Road, Croydon, Surrey CR9 2RU (01-684 4025).

FARNELL

Although not at the Show, the Data Precision 935 DMM which is being distributed by Farnell is well worth including in this review. The 935 is a high performance low cost meter designed for field use with all the range and function switches mounted to one side to allow single handed operation leaving the other hand free to use the probe.

With a $3\frac{1}{2}$ digit 0.5in. liquid crystal display, the 935 has 29 ranges for d.c. or a.c. voltage/current and a high (2.8V) and low (250mV) resistance range.



Calibration is guaranteed for one year and full protection for over voltage, overcurrent and high voltage transients is provided. All resistance ranges will tolerate 500V r.m.s. or d.c. without damage or loss of accuracy. A standard PP3 battery will power the 935 for over 200 hrs or an a.c. adaptor is available as an optional extra.

The £99 plus VAT price includes test leads, battery, instruction manual and a one year guarantee.

Farnell International Instruments Ltd., Sandbeck Way, Wetherby, West Yorkshire LS22 4DH (0937 63541).



CANADA

New frontiers are available to the Anik-B, the communications satellite put into orbit by the National Aeronautics Space Administration of America on behalf of neighbour Canada. The launch in December by a Delta 3914 was a first for the new self-correcting guidance system. The satellite weighs about a ton and is in a geostationary orbit. The altitude is 22,300 miles at a longitude of 109°W. The satellite can reach the whole of Canada from coast to coast and from the United States border to 75°N latitude. Beyond this the curvature of the Earth limits the usefulness of the vehicle. It brings Resolute on Cornwallis Island within the area covered. Resolute is about 210 miles East of the north magnetic

The satellite was built for Canada by RCA-Astro-Electronics and is reported to be the most powerful satellite for domestic use of telecommunications. Anik-B has 12 channels in the 6-4GHz band and six in the more powerful band of 14-12GHz. Telesat, which is the corporation that operates the Canadian communication satellite system, paid 20M dollars for the satellite itself and another 20M dollars for the launch facilities.

The name Anik has been used for the whole series of satellites, Anik-A in 1972 and another in 1973 have been extremely successful. The word Anik is Eskimo for brother and is aptly applied, for the system has brought together all the scattered places of the second largest country in the world. With the advent of Anik-B, specific targets can be reached, even urban areas can be included without

danger of interference. The terminal receivers can use smaller aerials (which means also cheaper) than those that were required for the first two satellites.

The channels are divided between the Canadian Department of Communications and the Canadian Broadcasting Company. The Department of Communications has leased four high frequency channels in order to test a number of educational and social service schemes. These will be beamed towards professional organisations in the cities as well as to remote areas. It seems that Canada is to implement the most ambitious and advanced programme ever attempted by the use of satellites. Health and Education, which includes the extension of such services to the Eskimo settlements in the far north, will include also the teaching of English as a second language.

The launch of Anik-B was witnessed by a contingent from the People's Republic of China. The contingent, including 35 trainee astronauts, came to inspect the launch facilities. They have already begun negotiations for a satellite communications system which would be put into orbit by shuttle in 1980. The Canadian Minister for Communications. Jeanne Suave, examined the launch facilities but admitted that she was also considering the European Space Agency's "Ariane" rocket launcher for future payloads as well as the shuttle facilities. This will probably depend largely on price requirements.

MORE ABOUT VENUS

The unique Venus expedition with its multiple vehicles is now answering some questions, but posing others. Some of the questions were answered before the probes entered the Venusian atmosphere. A number of questions were to the point before the mission, and these were based on previous missions, both American and Russian. The amount of scientific covering was extensive when the two Nations entered the vicinity of the planet with almost simultaneous missions. Much data had been received from the previous Russian landers and those from the USA. Some of those questions are answered and more will be in the months to come.

One important result of this mission was that the probes offered detailed information about the atmosphere supplementing that already available from the earlier Russian missions. Venus has no magnetic field in the sense that the Earth has; what does exist is very small, if any at all. Yet the effect of the solar wind is similar to that on the Earth. The bow shock wave which is set up in the Earth's atmosphere when the solar wind presses on the magnetosphere, appeared also in the atmosphere on Venus. This occurred

at some 750 miles above the surface of the planet. It is clear that the solar wind induces a magnetic field in the ionosphere. The shock area is the barrier to the electrons and protons of the solar wind. The strong magnetic field was apparent at the height of 750 miles. The energy involved by the leakage of some of the particles beyond the shock wave would be sufficient to heat up the atmosphere. The temperatures involved could reach 5,000°K.

The Orbiter spacecraft verified previous suggestions that the level of the ionopause varies in height according to the speed of the solar wind. The first week of observations by the Orbiter indicated a falling speed of the solar wind from 300 miles/second to 155 miles/second. The consequence of this was that the ionopause rose from 155 miles to 930 miles. This sort of change means that the level of the ionopause is very sensitive, so much so that if there is an eruption such as a flare on the Sun then the conditions in the atmosphere would very rapidly change.

It so happened that during the first week of the orbiter monitoring there was a flare. This resulted in an increase of the solar wind speed to nearly 400 miles/second. The ionopause was flattened and was only 155 miles above the surface of the planet. The temperature of the atmosphere has an interesting gradient. It was measured to be 27°C at 155 miles down to -90°C at 60 miles then increasing to 450°C at the surface.

In spite of the high temperature the presence of swamps and vegetation beloved of science fiction is not possible, for it would seem that most of any water that may have existed at an earlier date has now gone. This is deduced from the fact that very little hydrogen is now escaping, which would indicate that the loss of water has been completed. However, there is evidence that there was water in abundance but that it circulated in the atmosphere and was, by the action of ultraviolet radiation, disassociated and lost from the planet.

It is a peculiar world from the human point of view. It is not possible to see the Sun or even be sure where it is in the sky (sic). Visibility on the surface is about 1.8 miles but everything is red in colour. For the most part it would appear to be almost flat everywhere but at one place there was revealed a strip about 75 miles wide which showed an elevation of 10,000 feet. There is a tremendous amount of dust. This was revealed by the toppling of one of the probes which sent up a cloud of dust which took some minutes to settle. On the whole the terrain so far seems to be like Earth and not at all like the Moon, Mercury or Mars.

SKYLAB

The problem of the demise of Skylab has occupied space engineers for some time now. The dangers were headlined far and wide in spite of the factual reports from NASA. The space experts have been very conscious of the hazards. The space station weighs some 85 tons and in burning up during the final descent there could be about 400 or 500 pieces. Most of them would provide no more than a very magnificent firework display. There is then left the major hazard of two possibly large pieces that would survive. These could weigh around two tons.

To reduce this hazard mission con-

trollers have adjusted the station's orbit so that there is more control over the path of re-entry when the atmospheric drag takes effect. To this end a signal was sent to Skylab to command a change in attitude. This attitude will ensure that the solar panels will continue to absorb power from the Sun. An official stated that there will be enough electrical power to make some control possible down to the last moment. This control could be in action for perhaps two orbits. By firing the small thrusters it may be possible to ensure that it is a "splash down" and thus avoid damage to property. The major part of the deoris will be harmless, it is the large pieces that are cause for concern. Since they are potentially dangerous should they fall on land that is inhabited, every effort will be made to avoid this.

GROWTH OF SATELLITE COMMUNICATIONS

At the end of 1968 revenues from satellite communication services had reached some 420M dollars. The growth rate is such that the US market research organisation, Creative Strategies International, has forecast that the revenue by 1983 will reach one billion dollars. This is only part of a number of forecasts in the journal Satellite Communications.

ews Briefs

COUNTDOWN

Third International Symposium and Technical Exhibition of Electromagnetic Compatibility-May 1-3. Rotterdam. Details: Mr. R. E. Gerritson (070) 906 800.

Compec Europe—May 10. Centre International Rogier, Brussels. Details: 01-261-8437/8

Welsh Amateur Radio, TV, Electronics and Computer Exhibition .-Sunday, May 20. Barry Memorial Hall. There will be trade, and bringand-buy stands. Further details: Reg Rowles on 0222-565656.

Midlands Breadboard-May 23-26. Bingley Hall, Birmingham. Details: Trident.

Intel Fair-June 11, 1979. Wembley Conference Centre. Details to be

Transducer 79 + Testmex 79-June 19-21. Wembley Conference Centre. Details: Trident.

Great British Electronics Bazaar-June 28-29, 1979. Alexandra Palace, London. Details: 0799 22612.

1979 Microcomputer Show (incorporating the DIY Computer Fair) — July 5-7. Bloomsbury Centre Hotel. London. Will include seminars. Further details: Online Conferences Ltd. Tel: Uxbridge (0895) 39262.

Harrogate International Festival of Sound—August 18-19 (public), August 20-21 (trade) 1979. The Exhibition Centre + hotels. Details: Exhibition and Conference Services Ltd., Tel. 0423 62677.

Telecom '79-Sept. 20-26. Palais des Expositions, Geneve. Details: Secretariat Telecom '79, Orgexpo, 18 Quai Ernest-Ansermet, Case Postale 65, CH-1211, Geneve 4 (Suisse).

Eltro Hobby '79—October 3-7. Killesberg Exhibition Grounds, Stuttgart. Details: 01-236-0911.

Compec-November 6-8, 1979. Grand Hall Olympia, London. Details: Iliffe Promotions Ltd. Tel: 01-261-8437/8

Electronics 79—November 20-23. Olympia, London. Details: 021 705

Breadboard 79-December 4-8. Royal Horticultural Halls, Westminster. Details: Trident International Exhibitions. Tel. 0822

The next All-Electronics show will take place over the period April 29-May 1, 1980. Grosvenor House will again be the venue. Organisers are The Evan Steadman Communication Group. Details: 0799-22612.

HIGHER SPEED. LOWER POWER TTL

WO new third generation series of computer logic i.c.s have been announced by Texas Instruments Incorporated. They are advanced Schottley series SN74AS and advanced low power Schottley series SN74ALS. The high speed advanced series are designed for computer applications and systems that achieve efficiency through high speed switching capability. The SN74ALS series are intended for minicomputers, terminals and other electronic office equipment in which lower power consumption rather than high speeds improves efficiency.

Series SN74AS gates will be offered in 20 and 24 pin d.i.p. packages the latter with MSI functions. Series SN74ALS gates will appear in 14, 16, 20 and 24 pin d.i.p.

Initially, the high speed circuits will combine high performance gate and MSI arithmetic function. Twenty five device types will spearhead the low speed series. Included are gates and dual D and J-K flip-flops as well as MSI circuits.

HAMS RALLY

THE North Midlands Mobile Rally organised by The Midland Amateur Radio Society and Stoke On Trent Amateur Radio Society will take place on Sunday April 29th 1979 at Drayton Manor Park near Tamworth, Staffs. The Park is located on the A4091, which is within easy reach of the M1, M5 and M6 motorways, and is well signposted.

Visitors will be made very welcome. The rally opens at 11.30 a.m. Radio talk in stations on 2 metres and 70 centimetres.

Further details, car stickers etc. are available free on request from: Norman Gutteridge G8BHE, 68 Max Road, Quinton, Birmingham B32

POINTS ARISING

MICROPRINTER (February 1979)

The stripboard layout shows the correct pinouts for the BC212 transistors. The inset is wrong. Some links are missing from the stripboard, and these are: IC3 pin 4 to pin 9. IC7 pin 10 to +5V. IC3 pin 13 to IC5 pin 5. Junction at emitter of TR15 to +5V. The first link out from pin 13 on IC4 (the "bendy" one) should be removed, to prevent IC4 pin 13 from connecting to IC1 pin 13.

The value of "R" should be $10\Omega/\frac{1}{2}$ W.

SEQUENCER (April 1979)

For continuity between both sides of the main p.c.b. it is necessary to solder the component leads on each side and to use Soldercon pins for the i.c.s.

Michael Tooley B.A. David Whitfield B.A.

*HE sound operated switch described in this article can be used for any application where the presence of an audio frequency signal is required to provide control of an external circuit. A typical use is that of providing for automatic operation of transmitter receivers, or the start/stop control of tape recorders. In the former application the sound operated switch is inserted between the microphone and "press-totalk" rail, so that the operator's voice actuates the changeover relays. In the latter, the sound operated switch is used to provide automatic start of the tape recorder when the required input signal, such as speech, music, or even bird song, is present!

The block diagram of this system is shown in Fig. 2.

CIRCUIT DESCRIPTION

The complete circuit of the sound operated switch is shown in Fig. 1. The incoming audio frequency signal is applied to a conventional voltage amplifier stage, IC1d, which has variable gain between 1 and approximately 50 by means of VR1. The output of this first stage is nominally



several hundred millivolts peak to peak, and this is then fed to a precision unity gain active fullwave rectifier formed by D1, D2, IC1a, and IC1c. The positive output excursion of this rectifier is applied to a comparator, IC1b, the threshold of which is made variable by means of VR2. The negative output excursion of the comparator is used to enable a conventional timer circuit which uses a 555. The time constant of the timer monostable is controlled by means of VR3, and the positive going output of the timer is made available for control of external circuitry. A light emitting diode, D4, is incorporated in order to provide a visual indication of the output stage.

It is important to note that, once the input level has risen sufficiently to trigger the circuit and enable the timer, the output of the monostable will remain "high" (for a period of time dependent on VR3, R12, and C6) regardless of any subsequent reduction of input level.

In practice this means that there is a pre-determined delay in the output reverting to its original state and hence the circuit does not "drop out" at regular intervals when the input level falls during momentary pauses in speech or music. Those familiar with voice operated equipment will know that there is nothing worse than a voice operated switch that drops out at the end of each syllable!

CONSTRUCTION

With the exception of the sockets, I.e.d.s, and VR3, all the components are mounted on a small single sided printed circuit board. The layout of the p.c.b., viewed from the copper foil side, is shown in Fig. 3. The corresponding component layout, on the top side of the board, is shown in Fig. 4. Care should be taken to ensure that all components are located correctly on the p.c.b. and the use of sockets for the two integrated circuits is recommended. The p.c.b. is mounted using four short 6BA pillars and, by virtue of its relatively small size, it may be easily fitted into almost any vacant space inside an existing piece of equipment. Alternatively, where the circuit is to be used externally, the use of a standard die-cast aluminium or plastic box is recommended. In this case, the box should be sufficiently large to accommodate an internal battery.

TESTING AND INITIAL ADJUSTMENT

Carefully check the p.c.b. and external wiring before connecting the supply. Connect a 9V battery (PP3, PP6 or similar), or alternatively a stabilised d.c. supply of 8V to 12V, perhaps from the equipment to which the circuit is fitted. Insert a milliammeter on the 100mA range in the positive

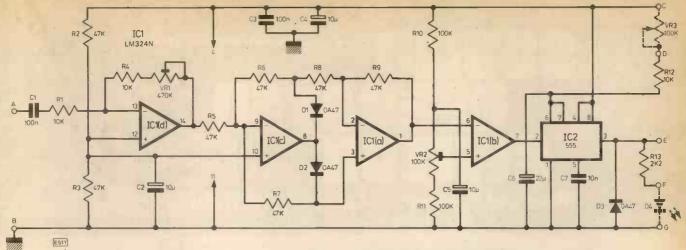


Fig. 1. Circuit diagram

SPECIFICATION

Maximum input sensitivity: 2mV pp sine wave at 1kHz

into $10k\Omega$

Frequency response: 60Hz to 15kHz at -6dB

(see Fig. 5)

Output control voltage: standby = 0V

triggered = 7.5V

Time delay (on reverting from triggered to standby): continuously variable from 200ms to 2.8s*

Attack time (on changing from standby to triggered):

typically less than 500µs

Supply: 9V d.c, at 7.5mA standby, and 11mA triggered.
Alternatively an external regulated supply of

between 8V and 12V may be used.

* Longer time delays may be obtained by increasing the value of VR3. A $1M\Omega$ variable will, for example, give a time delay of approximately 30s.

Fig. 2. Block diagram of Sound Operated Switch

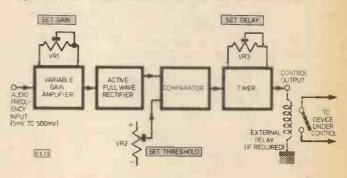


Fig. 3. Printed circuit board track layout (actual size)

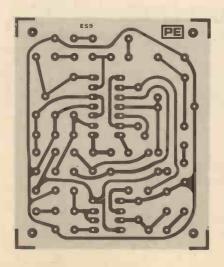
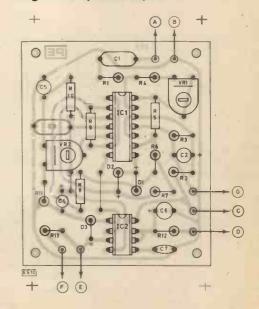


Fig. 4. Component layout



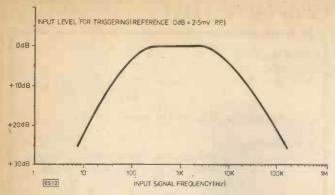


Fig. 5. Frequency response of the sound operated switch

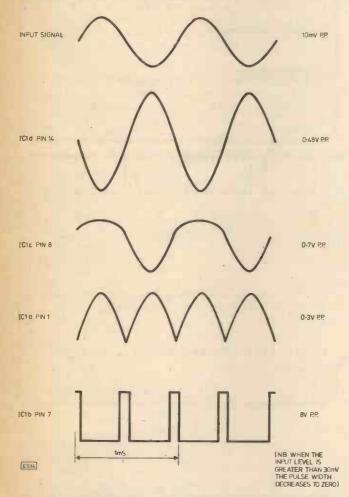


Fig. 6. Waveforms (input =10mV pp at 1kHz). When the input level exceeds 30mV, the pulse width on pin 7 decreases to zero

lead and check that the supply current is in the range 10mA to 20mA. Where an I.e.d. is incorporated to indicate that the supply is turned on, check also that this is illuminated. At one extreme setting of VR2, D4 should not be illuminated and at the other it should be illuminated. Carefully adjust VR2 until D4 is just extinguished. This should occur at about mid-position of VR2.

Set VR1 to maximum position (this corresponds to maximum gain) and connect a microphone or signal generator to the input. It should now be possible to trigger the circuit with an audio frequency signal of a few millivolts peak to peak.

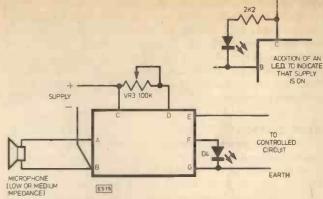
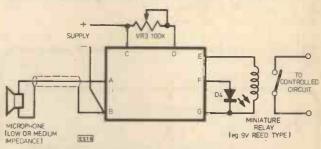
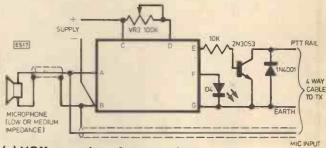


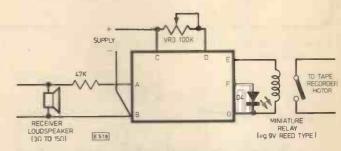
Fig. 7. (a) Basic arrangement of external circuitry



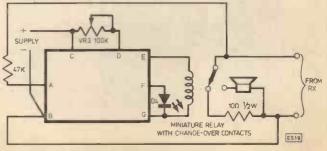
(b) Use of a relay to control an external circuit



(c) VOX operation of a transmitter



(d) Automatic recording system (audio connection to tape recorder not shown)



(e) Audio derived squelch for a receiver

VR2 can, if necessary, be re-adjusted for an increase in threshold sensitivity. Adjustment of VR3 should provide a variable delay in re-setting the output once the input signal has been removed. At minimum setting, VR3 should produce a delay which is hardly noticeable. At maximum setting the delay should be approximately three seconds. The state of the output of the switch is, of course, readily indicated by means of D4.

APPLICATIONS

The basic arrangement of the external circuitry of the sound operated switch is shown in Fig. 7(a). The microphone used should be a low or medium impedance type (300 Ω to 50k Ω). An I.e.d. may also be added, as shown, in order to give an indication that the supply is on. For many applications the use of a miniature relay is recommended, as shown in Fig. 7(b). A 9V reed type should prove suitable. When long microphone cables are necessary, the use of screened cable is recommended. This helps prevent spurious triggering of the switch due to hum and stray transients, especially when the circuit is used with maximum gain.

	-			-		
- 7/	A I	0.0	0	Lo	ы	6
w	w	tag	10	10	w	

		"standby"	"triggered"
IC1	pin 1	4.5	6.8
	pin 2	4.5	5.1
	pin 3	4.5	5.3
	pin 4	9.0	9.0
	pin 5	4.5	4.5
	pin 6	4.5	6.8
	pin 7	7.4	0
	pin 8	4.5	4.0
	pin 9	4.5	4.5
	pin 10	4.5	4.5
	pin 11	0	0
	pin 12	4.5	4.5
	pin 13	4.5	4.5
	pin 14	4.5	4.5

COMPONENTS LIST . . .

Resistors

R1, R4, R12 10k (3 off) R2, R3, R5–R9 47k (7 off) R10, R11 100k (2 off) R13 2k2

All fixed resistors 1W 5% carbon types

Potentiometers

 $\begin{array}{ll} \text{VR1} & 470 k\Omega \text{ subminiature horizontal skeleton preset} \\ \text{VR2} & 100 k\Omega \text{ subminiature horizontal skeleton preset} \\ \text{VR3} & 100 k\Omega \text{ linear carbon potentiometer} \\ \end{array}$

Capacitors

C1, C3
100n polyester (2 off)
C2, C5
10μ miniature 6V electrolytic (2 off)
C4
10μ miniature 12V electrolytic
C6
22μ miniature 12V electrolytic
C7
10n ceramic

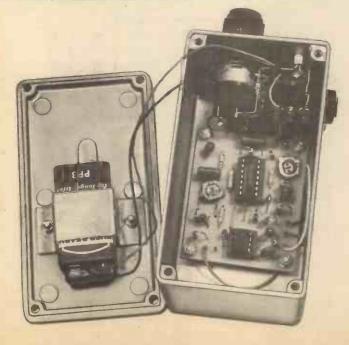
Semiconductors

IC1 LM324N quad operational amplifier IC2 555 timer (8-pin pack)
D1-D3 OA47 general purpose diode (3 off)
D4 I.e.d. light emitting diode

IC2	pin	1	0	0
	pin	2	7.4	0
	pin		0	7.2
	pin	4	9.0	9.0
	pin	5	5.8	5.8
	pin	6	0	rising to 5.8
	pin	7	0	rising to 5.8
	pin	8	9.0	9.0

The above voltages were measured using a 10M $\!\Omega\!\!\!\Omega$ input resistance voltmeter. The supply voltage was 9V.

If debugging is necessary, use these test point voltages

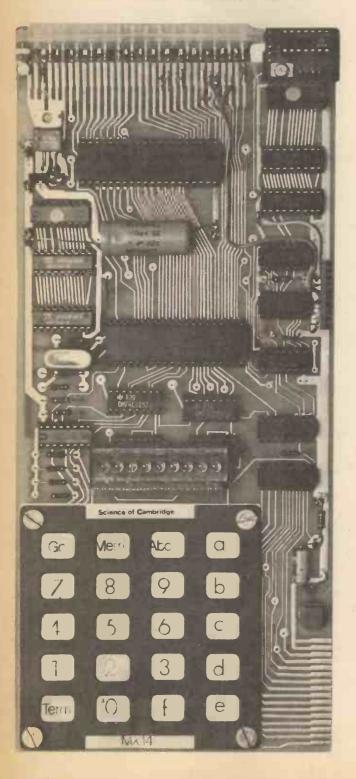


An arrangement for the VOX operation of a transmitter is shown in Fig. 7(c). The transistor is saturated whenever the switch is triggered. Thus the "press-to-talk" rail is effectively earthed whenever the operator speaks and this actuates the change-over relays in the transmitter or transceiver. An automatic recording system is shown in Fig. 7(d). In this application the presence of a signal at the loudspeaker causes the tape recorder motor to be energised. The recorder will run as long as signals are present. The $47k\Omega$ resistor connected in series with the input provides attenuation since, in this application, the sensitivity of the switch is too high and it is thus liable to become triggered by noise.

An audio derived "squelch" arrangement is shown in Fig. 7(e). When there is no signal from the receiver the loudspeaker output is terminated in a 10Ω resistor. The receiver is thus muted and background noise is eliminated. When a signal is present the loudspeaker is brought into circuit. The threshold of switching may be set by adjustment of VR1 and VR2.

NK14 REWIEW

A.A.BERK B.Sc. Ph.D.



The MK14 must surely represent one of the most remarkable revolutions in computer technology to date. Much has been said about this little kit, both in magazine articles by reviewers (see *Microbus* in Feb. issue) and word of mouth by constructors—whatever I or anyone else may say, the complete computer for around £40 is a landmark of, at present, unassailable proportions. The following, not uncritical, article pieces together my experience of the above device along with the description of the cassette interface by Science of Cambridge, and an interface to the *P.E. VDU* giving, all in all, a very cheap and powerful system.

THE kit comes pleasantly and well packaged as a set of resistors, capacitors, integrated circuits (without sockets), l.e.d. display, keyboard, reset switch, crystal, p.c.b. and even contains a 5 volt, 1 amp regulator. A "Micro Computer Training Manual" for the MK14 is also included.

The manual has clearly been designed with the absolute beginner in mind including, for instance, a good section on how to solder. However, it is generally agreed that absolute beginners and even beginners in either software or hardware alone, find it difficult to decipher some of the more fundamental concepts introduced in the manual.

The kit uses the National Semiconductors SC/MP II m.p.u. as its basis and the architecture and instruction set is described in the manual in just nine pages. Everything is there, but the beginner would do better to buy a full software introduction to SC/MP programming—I would recommend, for instance, Drury and Smart's "A Guide To SC/MP Programming".

MEMORY MAP

The hardware of the MK14 is also described in a cursory manner—even a complete memory map with addresses is lacking and for those who are contemplating using the MK14 the memory map shown opposite may be useful.

In addition to this omission, until recently, no pin numbers were included on the circuit diagram which made it very difficult to expand in any way—I am very pleased to see this rectified on the latest versions.

The upper four address lines of the system are left encoded (they have to be demultiplexed from the Data Bus anyway) and hence 12 BITs of address are available giving 4K of memory capability. This is the reason for the character "X" appearing in the memory map. The most significant hexadecimal character of the address must be used but may be anything—i.e. the addresses 0123 and F123, for instance, are indistinguishable.

The map indicates the hardware available in the kit. Included in the basic price of £39.95 + £3.20 VAT and postage etc., are 256 bytes of random access memory (using two 2111 chips). The display is a calculator type 9-digit, 7-segment multiplexed l.e.d. display giving an adequate read-out for simple programming. Space is provided on the p.c.b. for a further 256 bytes of RAM (two more 2111s) and an I/O device which itself contains 128 bytes of RAM.

If the above additions are made, the final system contains 640 bytes of RAM and 16 ports each of which may be separately configured as an input or an output line. "Handshaking" is also

XFFF -	
	STANDARD RAM
XF00	
	RAM I/O
XE00	101111111
AEUU	DICDI AV
	DISPLAY
XD00	
	RAM I/O
XC00	
	OPTIONAL RAM
XB00	OT HOWE KAIN
Abou	DAMI/O
	RAM I/O
XA00 -	
	DISPLAY
X900 —	
	RAM I/O
X800	
21000	MONITOR
32600	MONTOR
X600 —	MONTEMOR
	MONITOR
X 400 —	
	MONITOR
X200 -	
	MONITOR
X000	
71000	

Memory Map Table

possible via 8 of these lines for communication with an independent system of some kind—such as another MK14, as the manual informs us.

The memory map also shows how much of the 4K of immediate memory is lost by repeats. The 512 bytes of monitor, for instance, appears four times. This is due to incomplète memory decoding adopted to keep the cost to a minimum. However, it turns out that a small change is sufficient to free the extra monitor space and advice on this is given in the Issue IV kit.



The issue number, by the way, is found on the underneath of the p.c.b. This version explains how memory expansion can be effected to make a further $1\frac{1}{2}$ K of RAM available in the address locations X200 to X7FF (and not X2000 to X7FFF as incorrectly stated in the note). This would give a total of 2,176 bytes or RAM—quite respectable for the machine's size and basic cost, and excellent for the field of hardware control.

POOR FEATURES

There are two features which let the kit down, however. They are, firstly, the keyboard—which relies on bridging p.c.b. tracks with "conductive rubber"—and secondly, the extremely long lead times on delivery of the kit. I received one in January which was ordered in the previous Summer! Others have experienced waits from 2 to 4 months. The first of these criticisms is easy enough to overcome and the second is sure to ease in time—you might even try turning up at their doorstep or buying from a distributor to speed matters. The delay does not seem to extend to servicing, I only waited about fourteen days for them to check the monitor ROMs which I thought were faulty.

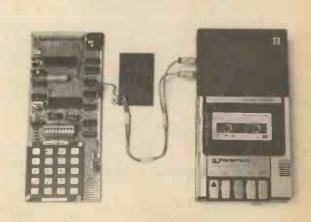
The issue IV board comes with a note to the effect that a kit of metal-domed switches may be purchased from Science of Cambridge for £2. The p.c.b. does, in fact, have an expansion edge connector next to the keyboard to add an external keypad, and though the manual gives no pin-out diagram for the connector, the tracks are easy to follow.

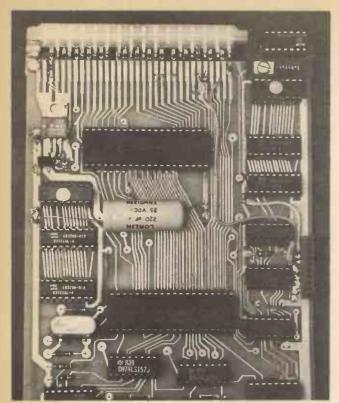
The kit is continually being updated. A few small mistakes in the earlier boards have been ironed out and the current p.c.b. is excellent—the only problem is in the extreme thinness of some of the tracks, and great care must be exercised not to break one of them.

One of the most important updates is the revised monitor program, available on fusible link PROM and pin compatible with the old one. This includes, among other things, an offset calculation for easier programming, single step facility (with some extra logic), and cassette tape interface control. The monitor occupies the same space as before by making some of the original routines more efficient.

CASSETTE INTERFACE

The photograph shows the cassette interface being used; and though slow (about four characters per second) it is very cheap, easy to use and of straightforward construction. The use of cassette files makes a tremendous difference to any machine, saving considerable amounts of time and energy. The interface uses two integrated circuits and some discrete components on a





The positioning of the connector and extra components to link with the P.E. VDU

small single sided p.c.b. The only fault I found was the marking of the positive end of the diode—wrong on the p.c.b. but correct on the circuit diagram. To store on cassette, the length and starting address must be set up in three memory locations. The length takes up a single byte and hence a maximum of 256 bytes may be stored at a time. The starting location of the cassette routine is then set up on the MK14's display and the recorder switched on. Pressing "GO" causes the block to be recorded—a l.e.d. on the cassette board flashes until the recording is complete. Taking information from the tape is a similar exercise. As the instructions implied, I found the volume control, on playback, to be very critical. The machine does work however, and several others I have talked to agree.

One of the most spectacular uses of the tape is in the direct loading of a block of information straight to a VDU screen. A set of notes can be kept on cassette and simply read back at leisure—especially if the tape has a tape counter for easy indexing. Another use would be for special screen formats stored on tape and used for particular programs to fill areas of the screen specially set aside for different functions. The programs involved would be very much simpler than normally necessary as the various lines and patterns used would not have to be generated in software—could be useful in the sort of minimal system for which the MK14 is ideally suited.

P.E. VDU INTERFACE

Of course a VDU or some sophisticated means of output, is essential for anything but the most basic uses of any machine, and below is described an interface to the *P.E. VDU* (published in October, November and December 1978).

The photograph shows that a 31-way Vero connector socket has been mounted on the top edge of the MK 14 p.c.b. This is not connected to the expansion edge connector in that region but is bussed, via wire links beneath the p.c.b. to the data, address and

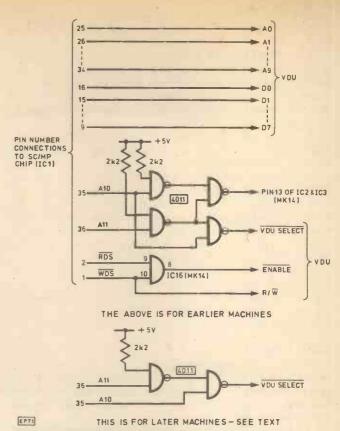
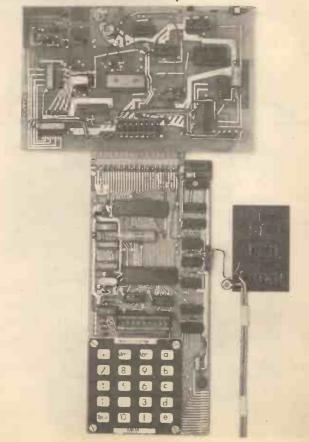


Fig. 1. Interface arrangements to the P.E. VDU

MK14 and P.E. VDU connected up



control lines of the SC/MP chip. In addition, a 4011 CMOS chip is mounted at top right of the p.c.b. and connected as shown in Fig. 1.

A spare gate in IC16 is used to enable the VDU only when RDS or WDS is high—this prevents the address and status information on the data bus from interfering with the VDU RAM. The circuit shown enables the VDU as a 1K block of memory replacing the top two copies of the monitor (each of 512 bytes).

To connect the interface to the MK14, "A11" must be disconnected from the enable inputs of IC2 and 3 (pin 13) and fed, along with "A10", to the 4011 as shown. Pin 13 of IC2 and 3 is then connected to an output of the 4011, again as in the diagram. The VDU "screen" then occupies the addresses X400 to X7FF. Later machines may have the monitor correctly decoded to lie only in the lowest 512 bytes—I have an issue V board which apparently has this modification. You can check by viewing a few addresses between X200 and X7FF to see if the monitor is there. If it is not then a simpler interface may be used which does not involve IC2 and 3. This is also shown in Fig. 1, the rest of the interface is as before.

It should be noted that no buffering was found necessary on any of the busses to drive the VDU as it presents, almost entirely, low power loads.

CONCLUSIONS

The expansions described will produce not only a reasonably sophisticated personal computer, but also a control development system of some power. In my opinion it is this latter area which has been somewhat neglected by Science of Cambridge's marketing philosophy.

In conclusion, the MK14 is a well thought out product nicely packaged, and easy to use with the revised monitor. Indeed, my firm opinion is that no electronics enthusiast or engineer should be without one in today's technology.



A reasonably sophisticated personal computer resulting from the use of the *P.E. VDU* with the MK14 and cassette interface

News Briefs

FINAL PHASE

N OUR January issue we gave a few details of the Lucas "How Far Can You Travel?" competition (see Electric Chariot Race). Well, the successful design stage entrants will now be battling it out on Sunday 2nd September 1979, at Donnington Park, Castle Donnington, Derby.

The competitor whose "home made" vehicle can transport him the furthest on two standard Lucas 12V batteries along the $\frac{1}{4}$ mile circuit, is the winner. Track marshalls will be on the lookout for barbed axles, and will "black flag" anyone seen slipstreaming behind another competitor.

FERRITE-CLAD AERIALS

A DISCOVERY by Professor J. R. James and Dr A. Henderson of the Electronics Department at the Royal Military College of Science, has resulted in the possibility of more compact whip aerials operating with HF and VHF band portable radios.

The discovery is that a thin cladding of ferrite material can be as effective as a thicker cladding of pure dielectric material in reducing the height of high frequency whip aerials, whilst offering the advantage of increased input impedance.

The ferrite may be of a coating formed in situ, or it can be achieved by sliding ferrite beads onto the conductor. The present lack of low-loss ferrites at high frequencies imposes an upper limit of about 100MHz generally, but the theoretical limit is considerably higher should new materials become available. Bulkiness limits the lower end to about 5MHz.

Manufacturing licensing arrangements for interested parties are being dealt with by Jim Strutt, Computers, Systems and Electronics Group, of the National Research Development Corporation, Victoria Street, London.

RAPID RECALL MOVE

ARGER premises at Wooburn Green now house Rapid Recall after a move from 9 Betterton Street, London.

The move was necessary because of the larger number of staff required to handle their franchise with the Digital Equipment Corporation, and cope with increasing sales of Intel and Intersil products.

The new headquarters are at: 6 Soho Mills, Wooburn Green, Bucks. Telephone, Bourne End (06285) 27072/5 and 24961. Telex 849439. Rapid Recall's northern office at Nantwich remains the same.

NASCOM CLUB

THE International Nascom Microcomputer Club is now officially launched with (at the time of writing) the publication of INMC News, issue one.

Further details about membership, which costs £5.00 per annum, are available from Tony Rundle at Nascom Microcomputers, 121 High Street, Berkhamsted, Herts.



WHEN attempting to record or amplify low level sounds (such as wildlife) with low output, low impedance microphones, you are often faced with the prospect of trying to pick out the desired sounds from under a blanket of hiss. This hiss or noise usually comes from the pre-amplifier. A good low noise pre-amplifier, as described here, can make all the difference between a foggy, messy sound that is hard to listen to, and a clean professional sound that is a pleasure to hear.

Having had a lot of trouble with noisy recordings, the author decided to put eight low impedance pre-amps through their paces to find the one with the lowest noise output. Two of the pre-amps used i.c.'s; and it was one of these that came top. It uses the ZN459 i.c. which is truly a low noise i.c.—the noise output voltage being a third of that for its transistor counterparts.

THE ZN459

The ZN459 is a versatile, wide band, low noise amplifier, operating up to 15MHz, with a voltage gain of 60dB. It basically consists of two amplifiers with gains of 100 and 10; giving 1000 (60dB) in all. The i.c. operates from a single supply of 6V maximum; has an input impedance of 7k at 10kHz, and internal input biasing. The i.c. lends itself to high performance circuits with few external components.

CIRCUIT DESCRIPTION

The circuit diagram of the Pre-Amp is shown in Fig. 1 and IC1 is operated from about 5.5V supplied from a 9V battery, via dropper R1, and held steady by C1. C2 decouples the input, and determines the low frequency limit. Bypass capacitor C3 prevents extraneous signals from reaching the input circuitry through the supply lines and causing instability. The output is taken via R2, and decoupling capacitor C4. R2 is to prevent excessive loading on the output. In determining the high frequency limit, it was found that RC frequency selective feedback caused instability, so

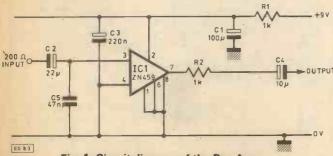


Fig. 1. Circuit diagram of the Pre-Amp

an alternative approach was used. C5 is connected straight across the input and effectively bypasses r.f. Fig. 2 shows the modifications necessary for a 600 ohm input. Pin 5 is a point halfway along the internal feedback path. By partially

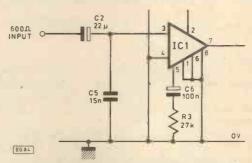


Fig. 2. Modification for a 600 ohm input

bypassing the a.c. feedback through C6 and R3 to 0V, the input impedance is raised to 600 ohms. C5 is changed to 15n to keep the same high frequency limit.

COMPONENTS ...

Resistors

R1, R2 1k (2 off) R3 27k

Capacitors

C1 100µ tant

C2 22µ tant

C3 220n tant

C4 10µ tant

C5 47n (200 ohm input) or 15n (600 ohm input) metallised polyester film

C6 100n tant

Semiconductors

IC1 ZN459

Miscellaneous

Input and output sockets

On/off switch

Metal case (80 x 55 x 25mm)

9V battery

Battery connector.

Holder for 8 pin i.c. (if req)

Veroboard 0-1in. matrix

CONSTRUCTION

EG 85

Construction is very straightforward and should present no problems. The Veroboard layout is shown in Fig. 3. Care should be taken to ensure correct polarity of capacitors, and

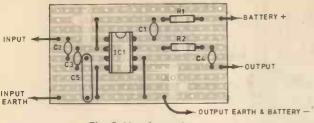


Fig. 3. Veroboard layout

that there are no accidental shorts between tracks. Put the i.c. in its socket last of all. The pre-amp should be housed in an all metal box for screening, with appropriate on/off switch, input and output sockets to suit individual require-

SPECIFICATION

 Input impedance
 200 ohm or 600 ohm

 Gain
 60dB

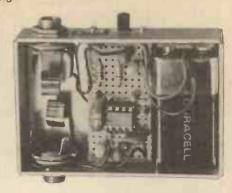
 Maximum Input Voltage
 1mV

 Frequency response
 25Hz — 35kHz + 0 - 3dB

 Distortion (THD)
 0.15% typ. (1V p-p, 10kHz)

Output Noise (unweighted) (Referred to an output level of 1V) 200 ohm

200 ohm 0.07mV (-83dB) 600 ohm 0.12mV (-76dB) ments. The circuit board should be mounted on non metal supports to avoid short circuits. The pre-amp is now ready for testing.



TESTING

Check there is about 5.5V across pin 2 and OV. Current consumption should be about 3mA. Plug in a microphone and monitoring source (with medium to high input impedance) and listen. The result should be a clean, distortion free, noise free sound.

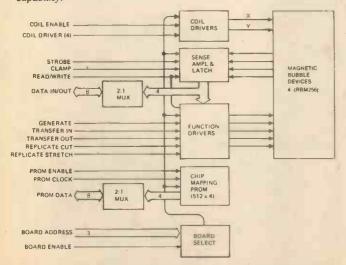
LIMITATIONS

With a maximum input of only 1mV before clipping occurs (due to the i.c.'s input circuitry), the pre-amp would be unsuitable for use with a singer's hand-held microphone, for example. Also, the pre-amp could be powered from existing equipment, but the supply would have to be exceptionally ripple-free, if not battery operated, to give good, hum-free results.

DEVELOPMENT SYSTEM

T LOOKS as though bubbles are going to continue hitting the news for some time. Another contender in this memory race is Rockwell, with a 256K-bit block access device. But to help encourage initial development towards bubbles by OEMs, Rockwell have produced a 4M-byte bubble development option for their System 65 microcomputer development system. It comprises one RCM650 controller with two RLM658 Linear Storage Modules.

The 6502 based development system can, with an auxiliary card cage, be expanded to the full 2M-byte bubble memory subsystem capability.



Linear bubble memory module (Rockwell). The chip mapping PROM holds redundancy information about faulty loops which may exist within the bubble memory.

News Briefs



55

AUTORANGING MUITINER PART 2

ALEX KOWALEWSKI Dr. MARK A. SAWICKI

This final part concludes with the remaining theory, and overall construction

DISPLAY BOARD

THE Autoranging Multimeter always requires that Over and Underrange must be detected. In our particular situation only digits 1 and 2 need to be sensed.

Overrange Digits 2, 3, 4 are all "off"

Underrange Digit 2 shows 0 or 1 and Digit 1 "off". Consequently segments 1b, 1c (connected in parallel) and 2a, b, f, d, must be interfaced so that the logic knows whether they are "on" or "off".

The complete Display Logic Interface employs five similar circuits as presented in Fig. 2.1 shaded portion.

INTERFACE CIRCUIT

If the segment is off, point A is at +5V, the transistor is off and B is -5V. If on, point A falls to approximately +3V due to the voltage drop across the l.e.d., the transistor turns on and point B rises to +5V. The capacitors on the interfaces for segments b and g are specially provided to stop false overrange indications during the 6 to 7 transition at digit 2.

The complete circuit diagram of the display logic interface is shown in Fig. 2.1. The decimal point and the lowest range l.e.d. indicator wiring diagram is quite straightforward.

Details of the p.c.b. layout can be found in a combined form at Fig. 2.3(a) and Fig. 2.3(b).

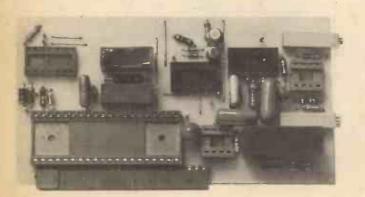
The display board and the Main Board are connected together using "edge inter-board connectors".

This is an economical and simple method of making rightangled connections between p.c.b.s by just inserting solder pins on the 0·1 inch p.c.b. pitch of the Display/Main Boards, i.e. plug (23-way) is soldered into the Display Board and the relevant socket is fixed to the Main Board. The DL 727 data is given in Table 2.1.

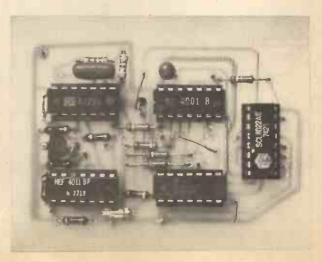
PSU

Approaching the final part of this project, we now give details of the power supply circuit. This is shown in Fig. 2.2, and the p.c.b. layout is shown in Fig. 2.4. The system uses a pair of popular LM309K regulators and all the other things are pretty standard. A protection fuse of 200mA on the

MAIN BOARD



LOGIC BOARD



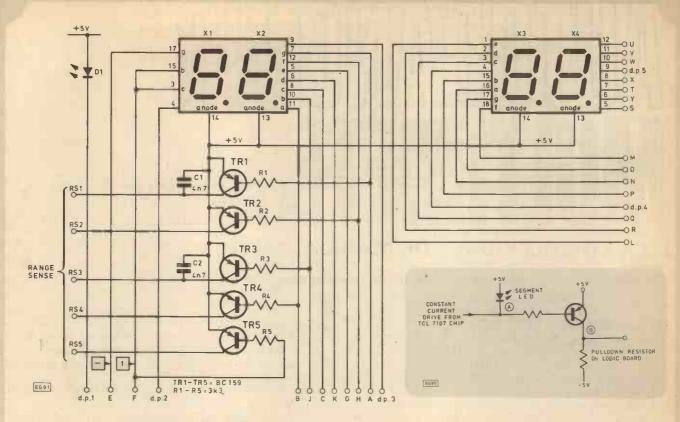
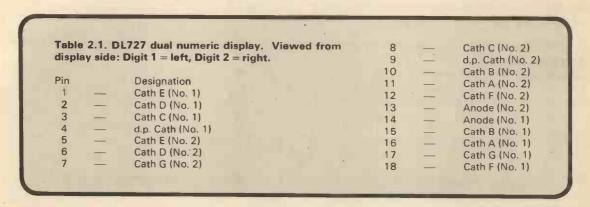
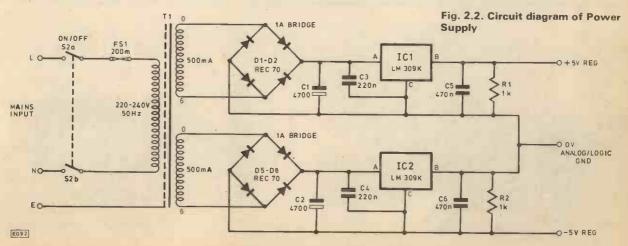


Fig. 2.1. Circuit diagram of Display Board





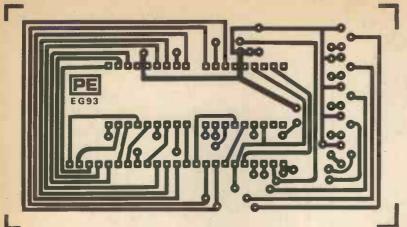
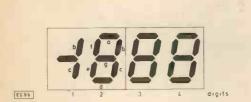
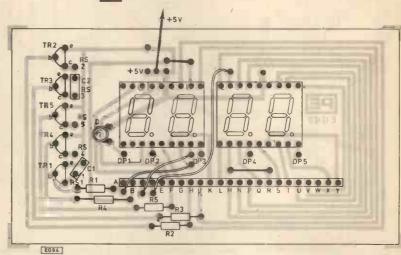


Fig. 2.3, (a) Display Board p.c.b.

Fig. 2.3. (b) Display Board component layout





COMPONENTS . . .

Resistors

R1-R5

3k3 (5 off)

4n7 (2 off)

Capacitors

C1, C2

Displays

Transistors

TR1-TR5

X1-X2, X3-X4

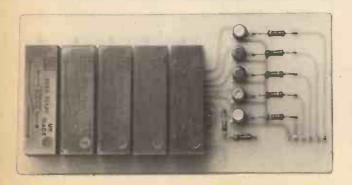
D1

BC159 (5 off)

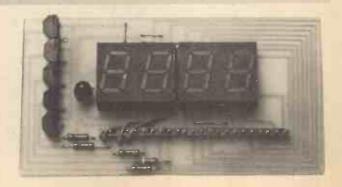
Litronix DL727 Dual 7-segment

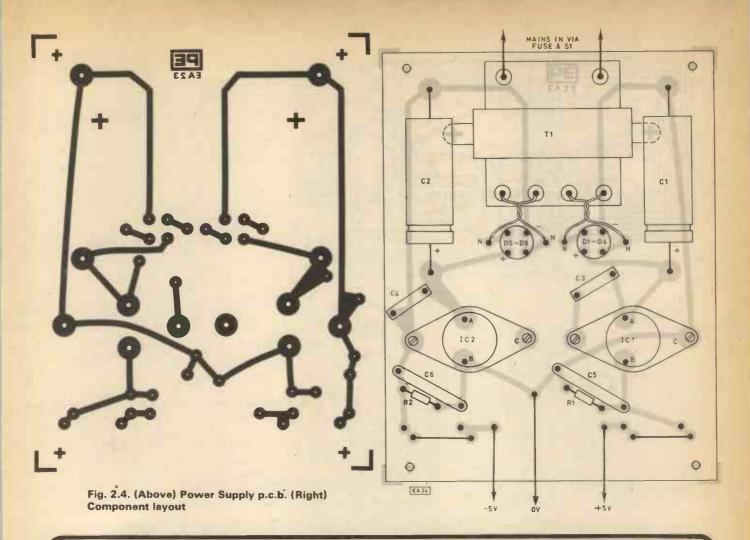
(2 off) 0.2 inch l.e.d.

RELAY BOARD



DISPLAY BOARD





COMPONENTS ...

Resistors

R1, R2 1k ¼W 5% (2 off)

Capacitors

C1, C2 4700µ/25V (2 off) C3, C4 220n (2 off) C5, C6 470n (2 off)

POWER SUPPLY

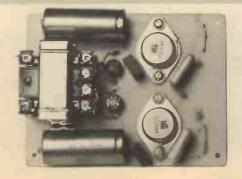
Semiconductors

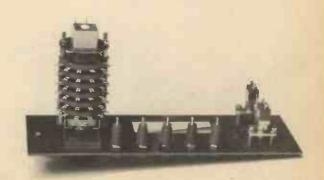
D1-D4, **D**5-D8 REC 70 (2 off) IC1, IC2 LM309K (2 off)

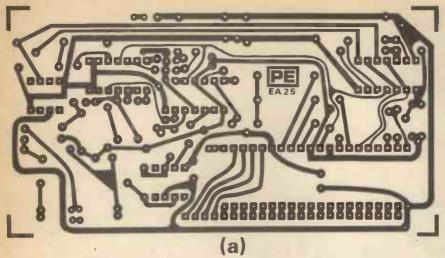
Transformer

1 240V/2×6V, 0·5A sec's. RS type 196–296 or equivalent (via Ace Mailtronix)

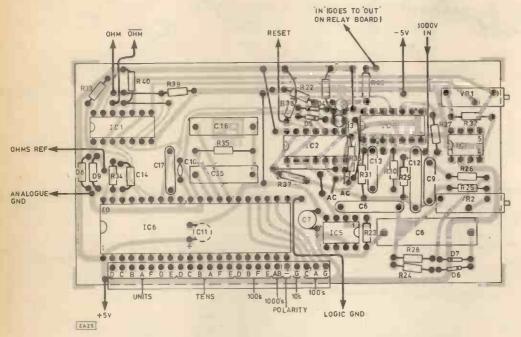
PSU BOARD







I/O pins	Control pin	Switch	Function
3-4 1-2 10-11 8-9	5 13 12 6	S3(b) S3(c) S4(a) S4(b)	OHM OHM OHM
3-4 1-2 8-9 10-11 10-11 1-2 3-4	5 13 6 12 12 13 5	S1(a) S2(a) S2(b) S3(a) IC3 S1(b) S1(c) S2(c)	AC AC AC AC AC



REF-02 Regulator

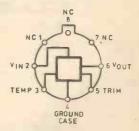
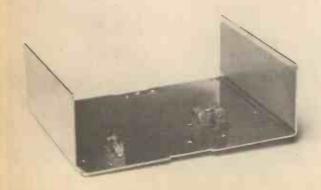


Fig. 2.5. (a) Main Board p.c.b. (b) Component layout

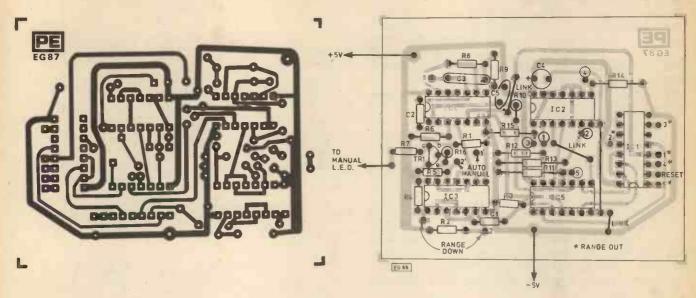
(b)





AUTORANGING MULTIMETER





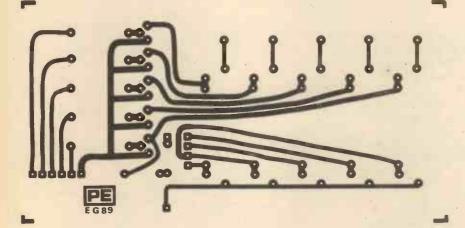


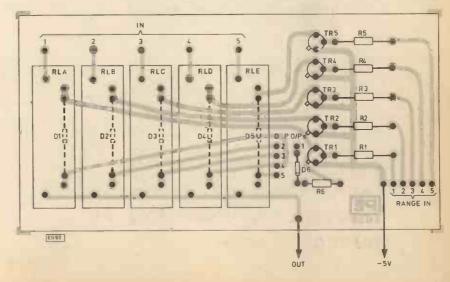
Fig. 2.6. (a) Logic Board p.c.b. (b) Component layout

Fig. 2.7. (a) Relay Board p.c.b. (b) Component layout

(a)

(b)

All p.c.b.s are shown full size. Dotted components are mounted on the p.c.b. track side. Some photographs differ from their associated layout diagrams where components have been mounted on the track side initially. It must be emphasized that track layouts are critical for stability, and must not be altered.



mains Live side follows two sets of contacts of the mains switch mounted right at the bottom of the function switch. In order to help you with the function switch wiring, Fig. 2.8 and Fig. 2.9 show both the top and the bottom halves of this switch's wafers and associated wiring.

MAIN BOARD PCB DETAILS

The Main Board p.c.b. layout is shown in combined form in Figs. 2.5(a) and 2.5(b). The reason for insertion of a few components on the copper side was to keep the layout around the 7107 device as similar as possible to Intersil recommendations. However, this does not create, in the author's opinion, any serious constructional problems.

Some additional information about component layout can be found from the photographs.

LOGIC BOARD

The Logic Board p.c.b. is shown in Fig. 2.6.

RELAY BOARD

The Relay Board p.c.b. layout is presented in Fig. 2.7(a) and 2.7(b). The only components soldered into the copper side are the five 1N4148 diodes in parallel to the reed relay coils. The reed relays chosen are the popular and economic, yet still high quality, RS348-986 type. The use of this particular type of relay is not critical as long as its rating and size are similar.

INSTRUMENT ENCLOSURE

Since most of you will be electronically minded rather than mechanically minded, you may prefer to get hold of a ready-made attractively styled instrument case complete with stick on feet. We used the RS standard case No. 509-901.

A display window was then cut in the front panel, and the rotary function switch, Mode/Auto manual and Range Down pushbutton switches with corresponding yellow Manual Led. were placed on the left side. All the input sockets (continental 4mm type) were set in at the bottom of the front panel. For ease of identification we used red, with the exception of the Ground socket, which was green.

All Autoranging Multimeter p.c.b.s and fuse holders (except the mains power supply fuse) are mounted into a specially provided aluminium chassis. Last but not least, the rotary function switch requires a little perspex plate (4mm thick) to lock the device and prevent overturning the switch.

CALIBRATION

After calibration of "Set Ref", described last month:

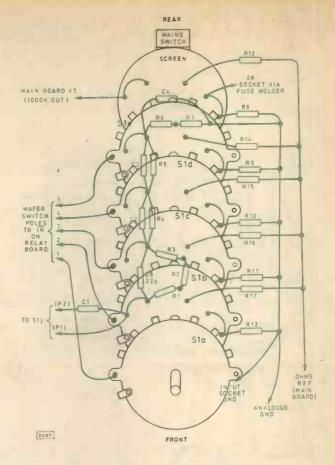
- If a known r.m.s. source is available or another accurate DMM, adjust AC Cal. until the reading agrees with the source or the other meter.
- (2) Short the 10n capacitor on function switch and C1, C2.

Preferably keeping to same range: put meter in DC Volts mode, measure any d.c. voltage. Switch to AC Volts mode and adjust AC Cal until reading is 3-2214 times the previous one in DC Mode.

Remove shorts.

IMPORTANT NOTE

It is advisable to calibrate the instrument without removing it from its grounded case.



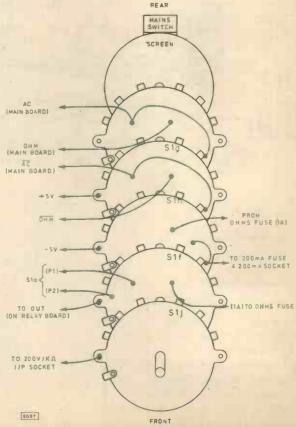


Fig. 2.8. (Top) Shows top half of S1 Fig. 2.9. (Above) Shows bottom half of S1

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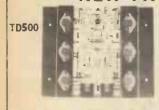
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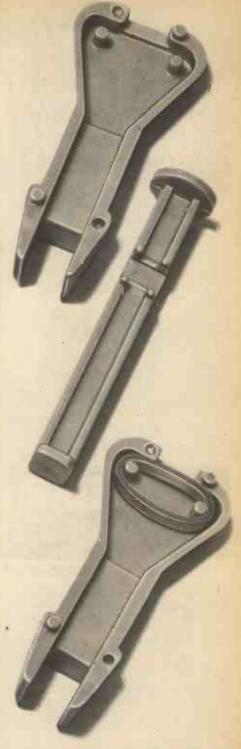


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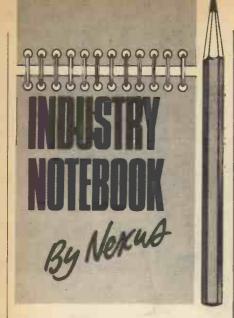
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The above photograph shows the tool dismantled. This enables the elastic band, which forms the plunger return spring, to be replaced should it ever perish. The two halves of the body can be parted by carefully inserting a blade in the joint. It is then a simple operation to replace the band and reassemble the tool.

The two halves are held by four fixed pegs which are a tight press fit, no adhesive is used on the joint. It is of course possible to use the tool without the return spring, though it will lose its smooth "feel".



Telecomms

Nothing illustrates better the cruel dilemmas resulting from electronics than the telecommunications business. The old faithful electro-mechanical telephone exchanges are having to give way to modern electronic exchanges requiring less space and less day-to-day maintenance as well as being more efficient and flexible in use and cheaper into the bargain.

But what an upheaval! The Strowger exchange using step-by-step switching through uniselectors and relays is labour-intensive in manufacture. Fine when you could employ low-cost female labour, displaced from the old textile mills, on assembly lines. But times have changed and not only are labour rates much higher today but legislation on equal pay means that the girls cost just as much as the lads. But in any case the run-down of manufacture of the old equipment means redundancy and we already have enough unemployment.

Plessey is one of the unhappy firms faced with problems of the gradual phasing out of Strowger production for the Post Office. Just to change over to electronic exchange production solves nothing because only a quarter of the workforce would be required. Either way people must go, especially as the Plessey factory in Liverpool is said to be losing £7 million per year. The Post Office is reported to be negotiating a revised price for Strowger equipment which will ease the financial problem for Plessey but not cure it in this particular plant. It is a problem which will continue well into the 1980s as electronics products continue their penetration of markets traditionally based on electromechanical devices.

The Strowger exchange has a curious history. Its inventor was not, as might be imagined, a highly trained scientist working in a well-equipped laboratory. Almon B. Strowger was a Chicago undertaker whose rival appeared to be getting all the orders for laying out, embalming and subsequent burial. The story goes that he attributed this not to his own lack of skill or even pricing policy but to manual switchboard girls who

directed all enquiries to the rival company. He thought an automatic switchboard not relying on people would ensure getting his share of undertaking work and so he set about inventing one. This was in 1889. By 1912 the Strowger system was in full use and, with only minor improvements, has continued to the present day and beyond.

For although the days of Strowger are clearly numbered the British Post Office will still be taking deliveries of the system until at least 1981 and assuming a service life of 20 years it seems that uniselectors will still be clicking away in the 21st Century. It just shows how a fundamentally sound system can last in the face of much more modern technology. The other example is the carbon granule microphone, virtually unchanged in a hundred years, still used in our telephones. Now the Post Office is searching for a modern electronics replacement, a unit which can be an exact replacement insert but with better and more consistent speech quality.

Time Scales

We are so frequently overwhelmed with a seemingly endless flow of new inventions and ideas that they seem almost instantaneous. Actually, time scales are often much longer than we imagine. The semiconductor effect was observed in lead sulphide and iron sulphide in 1874. It was 75 years later that the first crude point-contact transistor was made in a laboratory and another 13 years before the transistor found general acceptance in the electronics industry.

With all the present excitement about VLSI we tend to forget that the first MOS LSI circuits, albeit with a modest device count of only 2,500 elements per chip, were available ten years ago.

Some inventions have to wait for technology to catch up. Pulse code modulation (PCM) was proposed in the 1930s but had to wait for the invention of the transistor before becoming a practical proposition. The laser was a fine invention in 1960 but remained for years a scientific toy while waiting for suitable applications to be found.

On the other hand development can be quite rapid once the principles and the learning curve are established. In little more than a decade communications satellites have increased their capacity from 240 duplex voice channels (Intelsat II, 1967) to 12,000 channels (Intelsat V, 1979).

Speed

State-of-the-art in oscilloscopes was 500MHz bandwidth last year. Now Tektronix has made a huge leap forward in Model 7104 with a bandwidth of 1GHz, allowing direct observation of signals which on less sophisticated equipment are normally invisible. The writing speed of the trace is now an almost incredible 20cm/ns.

To get such performance Tektronix had to improve many aspects of oscilloscope design, not least the CRT itself. As it is virtually a microwave instrument, the p.c.b.s are based on microstrip techniques

and even the interconnections are based or microwave technology to reduce internal mismatches and reflection losses. Costing about £15,000 with an appropriate range of plug-in accessories, it is hardly designed for the impecunious hobbyist. But it will clearly appeal to the professional who can now record the fastest transients on ordinary film rather than the elaborate and cumbersome camera techniques formerly necessary.

Control Data Corporation's Cyber 203 computer, just announced, is also another record-breaker. Six times as fast as its predecessor the 203, using a method named vector processing, performs a handy 100 million calculations per second. The 203 is available in a range of sizes, the largest of which has a memory of two million 64-bit words. Price range is a little under £3 million to almost £6 million. First customer is the United States Air Force for three systems.

Computer World

Clearly, the big mainframe computers like the Cyber 203 still command a market but today's pioneering area is still in microsystems for the hobbyist, the small business and education. The United States, trend-setting as ever, now has hundreds of computer stores doing over-the-counter business. In the UK, Commodore is following suit with a dealer network for the Pet computer which is expected to expand to 100 retail outlets this year.

At the recent Microsystems '79 exhibition in London, some 60 firms showed systems costing as little as £100 up to £5,000 or so. Add-on sales are important. According to one market research report people are spending up to twice the initial processor cost on peripherals and other enhancements to their systems.

VLSI Stakes

GEC-Fairchild, having now appointed its full board of executives, settled on a site at Norton, Cheshire, and with appropriate finance lined up, look like meeting their target date for production next year of general purpose chips for the mass market, as much as two years ahead of Inmos, Britain's other new entrant. A strictly nononsense commercial venture, GEC-Fairchild while expecting to employ 1,100 people by mid-1981, will have all assembly work done in the Far. East where labour costs are lower. It may seem a pity but necessary if the project is to be successful.

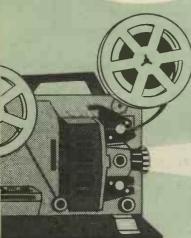
Incidentally, Fairchild has pulled out of the electronic watch business altogether and is cutting back on MPU-based video games. Market instability and low profit margins in the watch game are given as reasons. Fairchild made record sales in 1978, up to \$533 million with profits doubling to \$24.8 million.

Meanwhile, Inmos has been busy recruiting both in the UK and USA. The U.S. headquarters and technology centre are now to be at Colorado Springs. In the UK the Bristol site will be the technology centre but not necessarily the company HQ.

NEXT MONTH

SPECIAL WATCH OFFER!

We have been extremely careful with choice of watch—an alarm chrono, with stopwatch and dual time zone facilities. Price, we hope it will fall in the next month so it will be shown on the offer page only, to enable us to fix it at the last possible moment

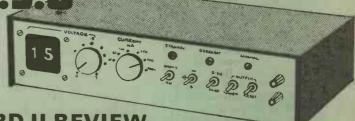


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WORKSHOP P.S.U

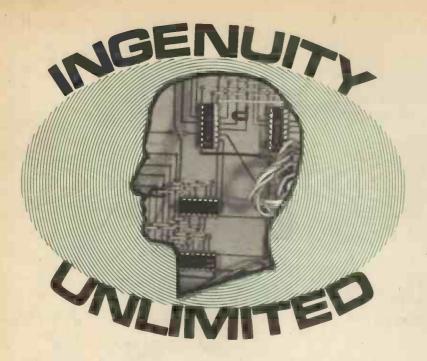
Regulated voltages in discrete steps up to 60V can be simply dialled using thumb-wheel switches. Output current is up to a maximum of 1A with switchable limiting. Overvoltage protection is also included.



PLUS OHIO SUPERBOARD II REVIEW

ELECTRONICS

OUR JUNE ISSUE WILL BE ON SALE FRIDAY, 11 MAY 1979



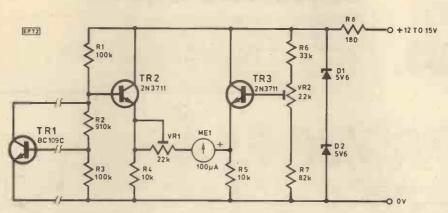
A selection of readers A selection of readers' original circuit ideas. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought. Why not submit your idea? Any idea published will be awarded payment

will be awarded payment according to its merits.

Articles submitted for publication should conform to the usual practices of this journal, e.g. with regard to abbreviations. tions and circuit symbols. Diagrams should be on separate sheets, not inserted in the text.

Each idea submitted must be accompanied by a declaration to the effect that it is the original work of the undersigned, and that it has not been accepted for publication elsewhere.

CHEAP THERMOMETER



THIS inexpensive thermometer gives suprisingly good stability and accuracy despite its simplicity.

The heart of the circuit is TR1, which forms the temperature sensitive probe with its own inbuilt amplification. This stage works in the V_{Inf} multiplier mode, common in power amplifiers, which amplifies its temperature dependent base-emitter voltage to about -25mV/°C. This voltage is then buffered by TR2.

VR2 picks off an offset voltage to bias TR3 such that the voltage between the emitters of TR2 and TR3 is proportional to the temperature. ME1 and VR1 form a voltmeter to measure this voltage.

TR2 and TR3 should be glued together for thermal tracking. The supply voltage is regulated by R8, D1 and D2. The use of a high voltage Zener diode is avoided because of its inherent temperature sensitivity. The metering unit is thus fairly free from temperature dependance enabling the probe alone to determine the meter deflec-

The probe should be connected to the metering unit via a suitable heat resistant cable. The leads of TR1 should be encapsulated in a silicon adhesive.

VR2 is adjusted so that the meter reads 0 with the probe in melting ice. VR1 is then adjusted so that the meter reads full scale with the probe in boiling water. The meter scale will then be linear, reading 0 to 100°C.

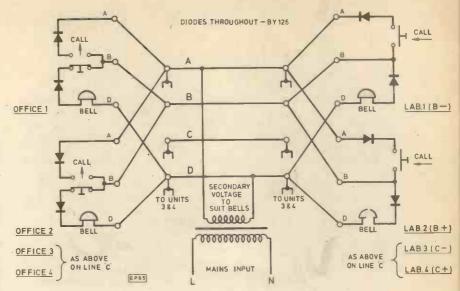
P. R. Williams, Stevenage, Herts.

4 WIRE-8 WAY BELL SYSTEM

WE have four offices, each associated with one of four laboratories some distance away. Each lab and office has a pair of phones on the same extension number, and my bell system is used mainly to call the phones on the other end of the 'pair'. The bell system is, of course, not electrically connected to the phones.

It can be seen that two of the four wires are the power lines and are fed (at any convenient point in the system—ours is fed in one of the labs) with an a.c. voltage to suit the chosen bells or buzzers. The other two calling lines are steered to the appropriate bells by diodes, thus each bell receives half wave rectified a.c. when its calling line is connected to the supply line via a diode of corresponding polarity.

Only two pairs of units are shown on the circuit diagram—the other two pairs are identical, but use calling line C. The office



units are shown with change-over push buttons—wired this way, the local bell does not ring when the push button is pressed. The lab units are shown with simple s.p.s.t. buttons—wired this way, each push would ring the bell at both ends of the line at once.

If line D is made an earth return, then the circuit becomes a 'three wire eight way' bell system!

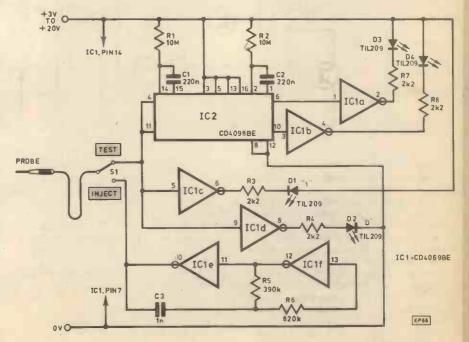
C. P. Finn, Southwood Park, Beverley.

CMOS LOGIC PROBE

THE probe employs two 'B' type CMOS integrated circuits to provide a comprehensive logic probe together with a square-wave signal injector. Four lightemitting diodes show the state of the circuit under test. D1 indicates a logic level '1', whilst D2 indicates a logic level '0'. These two are isolated from the external circuit via inverters.

Any logic pulses fire IC2, a dual monostable multivibrator. Positive pulses fire monostable 'a' which feeds D3, whilst negative pulses fire monostable 'b' which feeds D4. Thus a continuous wave-train will light all four. S1 switches between the logic test function and the signal inject function. The signal injector is formed from the two spare inverters in IC1, and oscillates at approximately 1kHz.

D. P. Akerman, Coventry.



VARITONE STYLOPHONE

HERE IC1 is connected as an oscillator with a frequency determined by the presets. By touching the probe onto the printed circuit board's "keys" (shown in the etching detail), which are connected to their corresponding presets, IC1 is made to give a frequency of between Bb7 (3729Hz) and G9 (12,544Hz). This squarewave signal is fed into IC2, a CMOS binary divider with its outputs connected to rotary switch S1, which selects the octave required and feeds the signal through TR2, via a volume control potentiometer. The amplified note now passes through a loudspeaker, giving a musical sound. Any number of keys may be included, providing there is a preset to tune each note.

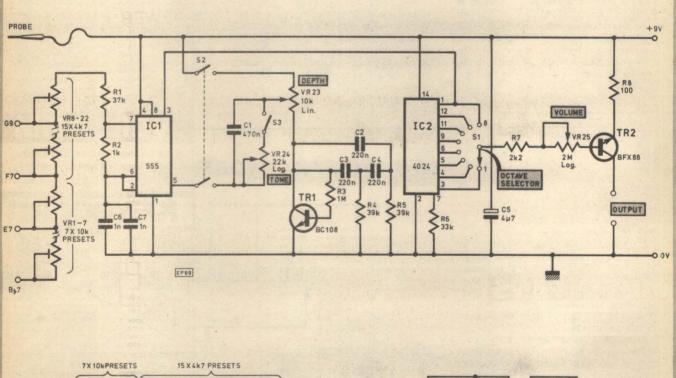
A tremolo effect is added using the circuit around TR1, which provides a 6Hz modulation of variable depth when fed into pin 5 of IC1. Some interesting effects may also be obtained using the "tone control" potentiometer. However, this does alter the pitch of the note produced, and so a switch is included to disconnect it in normal operation.

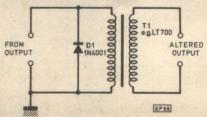
The squarewave tone output may be modified by passing it through an audio transformer T1 as shown, D1 being required to soak up any back e.m.f. from the windings. The modified output has a harsher quality than the plain squarewave sound.

The instrument is tuned using the fifteen 4.7 kilohm and seven 10 kilohm presets

connected in series. They should be adjusted to tune the notes starting at the highest and working down the scale.

T. J. Barker, Reigate, Surrey.







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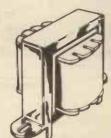
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7433 - 15° 74113 - 35° 74176 101° 74295 - 105° 7437 25° 25° 74114 - 35° 74177 101° 74298 - 125° 7438 25° 25° 74121 27° 74177 101° 74298 - 125° 7404 17° 10° 74123 50° 75° 74180 101° 29° 74366 - 51° 7441 70° 74123 50° 75° 74182 21° 29° 74366 - 51° 7442 50° 55° 74124 - 125° 74182 10° 74367 - 51° 7445 50° - 74125 51° 39° 74185 152° 74386 - 39° 7446 60° - 74128 51° 39° 74185 152° 74386 - 39° 7447 60° 87° 74133 - 19° 74190 317° 225° 74670 - 185° 7448 60° 67° 74133 - 19° 74190 121° 75°	4035 1.06* 4528 92* 711209 4040 92* 4534 7.12* 711209 4042 7.0* 4536 3.74* 711218 4043 81* 4543 1.62* 711218 4049 A3* 4556 1.51* 711232 405* 81* 4583 1.02* 405* 81* 4583 1.02*	Red x .15° .10° .10° .09° TIL220 Yel x .20° .18° .16° .14° TIL224 Yel x . Red x .20° .18° .16° .14° TIL228 Red x	10 + 50+ 100+



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APPLICATIONS: hi-fi; mixers; disco; guitar and organ; public address.

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HY30 15W into 8Ω

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APPLICATIONS: updating audio equipment; guitar practice amplifier; test amplifier; audio oscillator. SPECIFICATION: Output Power—15W R.M.S. into 8Ω. Distortion—0·1% at 15W. Input Sensitivity—500mV. Frequency Response—10Hz-16kHz -3dB.

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external components.

APPLICATIONS: medium power hi-fl systems: low power disco: guitar amplifier.

SPECIFICATIONS: medium power hi-fl systems: low power disco: guitar amplifier.

SPECIFICATION: Input Sensitivity—500mV. Output Power—25W R.M.S. into 80. Load Impedance—4-160. Distortion—0-04% at 25W at 1kHz. Signal/Noise Ratio—75dB. Frequency Response—10Hz-45kHz - 3dB. Supply Voltage—±25V. Slze—105 × 50 × 25mm.

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The HY120 is the baby of I.L.P.'s new high power range, designed to meet the most exacting requirements including load line and thermal protection this amplifler sets a new standard in modular design

FEATURES: very low distortion; Integral heatslink; load line protection; thermal protection; five connections; no external components.

APPLICATIONS: hi-fi, high quality disco-public address; monitor amplifler, guitar and organ.

SPECIFICATION: hi-fi, high quality disco-public address; monitor amplifler, guitar and organ.

SPECIFICATION: Input Sensitivity—500mV. Output Power—60W R.M.S. into 8Ω. Load Impedance—4-16Ω. Distortion—0-04% at 60W at 1kHz. Signal-Noise Ratio—90dB. Frequency Response—10Hz-45kHz—3dB. Supply Voltage—±35V. Size—114 x 50 x 65mm.

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FEATURES: thermal shutdown; very low distortion; load line protection, integral heatsink; no external

omponents.

APPLICATIONS: hi-fi: disco: monitor, power slave: industrial, public address.

APPLICATIONS: hi-fi: disco: monitor, power slave: industrial, public address.

SPECIFICATION: Input Sensitivity—500mV. Output Power—120W R. M.S. into 80. Load Impedance—

4-160. Distortion—0:05% at 100W at 1kHz. Signal/Noise Ratio—96dB. Frequency Response—10Hz—45kHz—3dB, Supply Voltage—±45V. Size—114 × 50 × 85mm.

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HY400 240W into 4Ω The HY400 is I.L.P.'s "Big Daddy" of the range producing 240W into $4\Omega!$ It has been designed for high power disco or public address applications. If the amplifier is to be used at continuous high power levels a cooling fan is recommended. The amplifier includes all the qualities of the rest of the family to lead the market as a true high power hi-fidelity power module.

FEATURES: thermal shutdown, very low distortion; load line protection; no external components. APPLICATIONS: public address; disco. power slave, industrial. SPECIFICATION: Output Power—240W R.M.S. into 4Ω. Load Impedance—4-16Ω. Distortion—0-1% at 240W at 1kHz. Signal/Noise Ratio—94dB. Frequency Response—10Hz-45kHz. –13dB. Supply Voltage. — ± 45V. Input Sensitivity—500mV. Size—114 × 100 × 85mm.

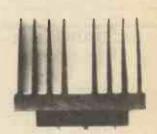
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AC141K	0.35	BD137 BD138	0.35	OA79	0.30	1N5401	0.13	7423 7425	0 32
AC142 AC142K AC176	0.20 0.30	BD13B	0.40	OAB1 OAB5	0.30	1S44 1S920	0.04	7425	0.30
AC176	0.20	BD140	0.44	OA90	0.08	1S921 2G301	0.07	742B 7430	0.43
AC187 AC188	0.20 0.20	BD144 BD181	2·00 1·10	OA91 OA95	0.08	2G301 2G302	1.00	7430	0-17
	0.20	BD182	1.18	0A200 0A202	0.09	2G306	1.10	7433	0.36
ACY18 ACY19 ACY20	0.80	BD237 BD238	0.40		1.00	2N404 2N696	1.00 0.25	7437 7438	0.32
ACY19	0.75	8DX10	0.91	OAZ200 OAZ201 OAZ206 OAZ207 OC16 OC20 OC22	1.00	2N697	0.25	7440	0.18
ACY21 ACY39 AD149	0.75	8DX10 BDX32 BDY20	2.00	OAZ201	1.00	2N698	0.30	7441AN	0-85
ACY39	1.50 0.70	BDY60	1.25	OAZ206	1.00	2N705 2N706	1·20 0·15	7442 7447AN	0.72
AD161	0.45	BF115 BF152	0 25	OC16	2.00	2N706 2N708	0·15 0·20	7450	0.18
AD162 AF106	0-45	BF152 BF153	0.18	0020	2.50	2N930	0.20 0.26	7451 7453	0·18 0·18
AF114	0 75	BF153 BF154	0-20 0-17	OC22 OC23 OC24 OC25 OC26 OC28 OC29	2.75	2N1131 2N1132	0.26	7454	0-18
AF115	0.75	BF159 BF160	0.16	OC24	3.00 0.90	2N1302 2N1303	0.35	7460 7470	0·18 0·35
AF116 AF117	0.75	BF167	0.20	OC26	0.90	2N1304	0.45	7472	0.33
AF139	0.40	BF173	0.20	OC28	2 00	2N1305	0-45	7473	0.36
AF186 AF239	1.20 0.45	BF177 BF178	0.24	OC35	2.00 1.50	2N1306 2N1307	0.50 0.50	7474 7475	0.40
AFZ11	2.75	BF179	0.25	OC35 OC36 OC41	1.50	2N1308	0.55	7476	0.40
AFZ12	2.75 0.40	BF180 8F181	0.30	OC41	0.80	2N1309 2N1613	0.55	7480 7482	0.55 0.75
AFZ11 AFZ12 ASY26 ASY27	0.40	BF182	0.30	OC42 OC43	2.25	2N1671	1.50	7483	0.90
A5Z15	1.25	BF183	0.25	OC44	0.60	2N1893	0-25 1-75	7484 7486	1.00 0.35
ASZ16 ASZ17	1.25 1.25	BF184 BF185	0.25	0C45	0.55 0.55	2N2147 2N2148	1 65	7490	0.52
ASZ20	1.50	°BF194	0.09	OC45 OC71 OC72	0.55	2N2218 2N2219	0.25	7491AN	0.80 0.60
ASZ21 *AU110	2.00 1.70	*BF195 *BF196	0.09	OC73 OC74 OC75	1.00 0.65	2N2220	0.18	7492 7493	0.60
°AU113	1.70	*BF197	0.12	OC75	0.65	2N2220 2N2221	0.18	7494	0.80
*AUY10 *BA145	1.70	BF200 *BF224	0-27	OC76 OC77	0.55	2N2222 2N2223 2N2368	0·18 2·75	7495 7496	0.72
*RA148	0·13 0·13	I REZAA	0.28	OC81	0.65	2N2368	0.17	7497	3.00
BA 154 BA 155	0.09	BF257 BF258 8F259	0.24	OC81Z OC82	0.65	2N2369A 2N2484	0·21 0·20	74100 74107	0.45
BA156	0.09	8F259	0.32	OC83	0.65	1 2 N 2 B 4 B	0.55	74109	0.70
BAW62	0.05	BF336 BF337	0.30	OC122	0.65	2N2904 2N2905	0.25 0.25	74110 74111	0.50
BAX13 BAX16	0.09	BF338	0-31	OC122 OC123 OC139	1.75	2N2906 2N2907	0.21	74116	1.75
BC107	0·12 0·12	8FS21 8FS28	3.96 2.23	00139	2.25	*2N2907	0.21	74118 74119	1.00
BC109	0-13	*BFS61 *8FS98	0-20	OC140 OC141	3.25	°2N2925	0.22	74120	0.83
*BC113	0.12	*8FS98 BFW10	0.20	OC170	1.00	°2N2926 2N3053	0.14	74121 74122	0.40
*BC115	0.14	REW/11	0.65	OC200	1.50	2N3053 2N3054 2N3055	0.50	74123 74125	1.00
BC108 BC109 *BC113 *BC114 *BC115 *BC116 *BC117 *BC118 BC125 *BC126	0.15	BFX84 BFX85 BFX87 BFX88	0·22 0·23	0C141 0C170 0C171 0C200 0C201 0C202 0C203 0C204	1.75	2N3055 2N3440	0.70	74125	0.55
*BC118	0.10	BFX87	0-21	OC203	1.75	2N3441	0.80	74128	0.60
BC125	0·16 0·20	BFX88 BFY50	0-21 0-26	00204	2.50	2N3442 2N3525	0.80	74132 74136	0.70
*BC126 *BC126 *BC135 *BC137 *BC147	0.14	BFY51 BFY52	0.26	OC205 OC206 OC207	2·50 2·50	2N3614	1.50	74141	0.80
*BC136	0-15 0-15	BFY52 BFY64	0-26 0-26	OC207 OCP71	1.75 1.25	*2N3702 *2N3703	0.11	74142 74143	2.30
*BC147	0.15	BEY90	1.25	ORP12	0.75	*2N3704	0.13	74144	2.50
BC148	0 08	BSX19 BSX20	0.21	*R2008B	1.75	*2N3705	0.13	74145	0-90 2-00
*8C149	0.09	BSX21	0 20	*R2009 *R2010B	2·25 1·75	*2N3706 *2N3707	0.13	74148	1.75
°BC158	0.08	BT106	1.25	TICAA	0.30 1.20	I 2N3708	0-10-	74150 74151	1.60 0.85
BC148 *BC149 *8C157 *BC158 *BC159 *8C167	0.10	BTY79/4 *BU205	3·19 1·75	T1C226D T1L209	0.20	2N3709 2N3710	0.10	74154	1.75
BC170	0.11	*BU205	2.25	*T1P29A *T1P30A	0 41	2N3711 2N3771	0-10 1-75	741 5 5 741 5 6	0.85 0.85
BC170 BC171 BC172 BC173	0-10	*BU208 BY100	0.45	T1P31A	0.45	2N3772	2.00	74157	0.75
BC173	0.12	BY100 BY126	014	T1P32A T1P33A	0 48	2N3773 2N3819	3.00 0.36	74159 74170	2.10
BC1//	0.15	BY127 BZX61	0.15	T1P34A	0.73	J 2N3820	0.45	74172	4-40
BC178 BC179	0.16	Series	0.13	T1P41A T1P42A	0.63	2N3823	0.55 0.72	74173 74174	1.40
BC182 BC183	0.11	BZY88 Series	0.13	T1P2955	0.67	*2N3866 *2N3904	0.13	74175	0 90
BC183 BC184	0.11	CRS 1/05	0.45	T1P3055	0.56	*2N3905 *2N3906	0.13	74176 74178	1.10
BC212 8C213	0.13	CRS 1/40 CRS 3/40 CRS 3/60	0.60	*T1S43 *ZS140	0.25	°2N4058	0.14	74179	1.25
8C213 8C214	0.15	CRS3/60	0.90	*ZS170 *ZS178	0·21 0·54	*2N4059 *2N4060	0.10	74180 74190	1.15
BC237	0.09	GEX66 GEX541 GJ3M	1.50	°ZS271	0.23	*2N4061	0.12	74190	1.50
BC237 BC238 BC301	0-25	GJ3M	0.75	*ZS278	0.57	*2N4062 *2N4124	0.13	74192	1.35 1.35
BC303 BC307 *BC308	0.24 0.10	GJ5M GL7M	0.75	*ZTX107 *ZTX10B	0.11	*2N4126	0.15	74193 74194	1.25
*BC308	0.10	GMO378A	1.75	*ZTX109	0.12	*2N4286	0.20	74195	1.00
*BC328	0.20	*KS100A MJE340	0-45 0-80	*ZTX109 *ZTX300 *ZTX301	0·12 0·13	*2N4288 *2N42B9	0.24	74196 74197	1-10
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°BC338	1.00	MJE371 MJE520	0.61	*ZTX303	0·17 0·19	*2N545B *2N5459	0.35	*76013N	2·25 1·75
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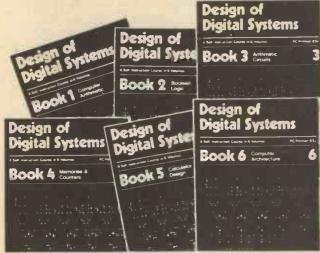
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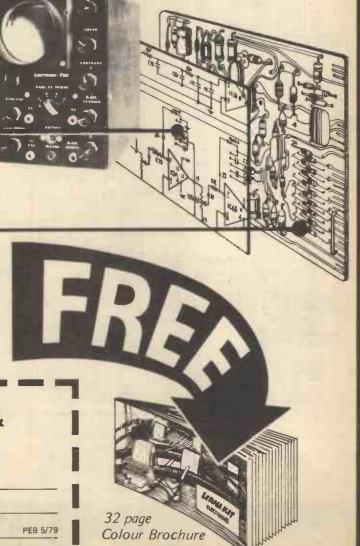
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AC126 0.17 BC109B 0.07 BC182A 0.09 BC541 AC127 0.16 BC109C 0.07 BC182B 0.09 BC541 AC127K 0.23 BC113 0.12 BC183A 0.09 BC541 BC541	0.11 BDY16 0.50 BFY19 0.50 A 0.11 BDY20 0.50 BFY33 0.50	OC203 2.50 2N2368 0.10 OC207 2.70 2N2484 0.18 OC1072 2.50 2N2904 0.18	7404 .13 7486 .26 74 7405 .13 7489 2.00 74	165 .90 4006 .95 4078 .20 166 1.00 4007 .20 4081 .20 167 2.70 4008 .94 4082 .20 170 1.70 4009 .46 4502 .90
AC128K 0.25 BC116 0.13 BC183LB 0.10 BC544 AC128/1760.42 BC117 0.15 BC184 0.08 BC548 BC548	A 0.11 BF123 0.20 BFY51 0.12 B 0.11 BF125 0.20 BFY52 0.12	OC1074 2.50 2N2906 0.18 R2008B 2.10 2N3053 0.15 TIP30 0.36 2N3117 1.00	7407 .29 7491 .65 74 7408 .14 7492 .45 74 7409 .14 7493 .36 74	172 4.00 4010 .50 4507 .52 173 1.20 4011 .15 4508 2.30 174 .90 4012 .15 4510 1.10
AC153 0.55 BC119 0.25 BC184LB 0.10 BC544 AC158 0.50 BC125 0.15 BC186 0.19 BC544 AC176 0.16 BC125B 0.16 BC187 0.19 BC554	A 0.11 BF137 0.21 BRY39 0.35 B 0.11 BF152 0.15 BSX20 0.18	TIP31 0.45 2N3440 0.70 TIP31A 0.45 2N3638 0.15 TIP32 0.46 2N3638A 0.16 TIP33 0.60 2N3643 0.24	7411 .18 7495 .55 74 7412 .21 7496 .62 74 7413 .25 7497 2.40 74	176 .90 4014 .85 4512 .95 177 .90 4015 .80 4516 .70 178 1.20 4016 .45 4518 .65
AC187 0.50 BC126 0.15 BC204 0.08 BC55: AC187K 0.55 BC136 0.15 BC204B 0.09 BC55: AC188K 0.55 BC137 0.15 BC206 0.10 BC55:	0.11 BF157 0.37 BSY20 0.21 A 0.11 BF158 0.15 BSY25 0.30	TIP42 0.60 2N3692 0.24 TIP2955 0.65 2N3705 0.06 TIS90 0.18 2N3706 0.06	7416 .27 74105 .40 74 7417 .27 74107 .28 74	179 1.10 4017 .55 4520 1.10 180 .70 4018 .90 181 1.95 4019 .45 182 .75 4020 .60
ACY20 1.02 8C138 0.30 8C208 0.11 8C55 ACY22 1.02 8C140 0.27 8C207 0.10 8C55 AD142 0.87 8C141 0.29 8C2098 0.09 8C55 AD143 0.87 8C142 0.20 8C212A 0.10 8C55	CC 0.11 BF173 0.20 BSY27 0.30 BSY29 0.54 A 0.11 BF177 0.25 BSY39 0.15	2N344B 0.30 2N3707 0.06 2N404 0.45 2N3711 0.06 2N524 0.48 2N3819 0.20	7421 .28 74110 .46 74 7423 .25 74111 .70 74 7425 .22 74116 1.60 74	184 1.20 4021 .85 185 1.20 4022 .85 186 7.20 4023 .20
AD149 0.65 BC147 0.06 BC213L 0.10 BC551 AD161 0.35 BC147B 0.07 BC214L 0.10 AD161/1620.70 BC148 0.06 BC237 0.15 BC751	B 0.11 BF179 0.25 BSY54 0.39 0.55 BF180 0.20 BSY56 0.36	2N526	7427 .25 74119 1.30 74 7428 .34 74120 .82 74	188 5.40 4024 ,68 190 1.10 4025 ,20 191 1.00 4026
AD162 0.36 BC1488 0.07 BC237A 0.16 BCY7 AD263 0.36 BC149 0.06 BC238 C.15 BCY7	0.13 BF182 0.20 BSY84 0.30 BU105 1.08	2N705 0.10 2N3906 0.06 2N706 0.10 2N4037 0.25 2N706A 0.11 2N4058 0.10 2N708 0.12 2N4222A 0.65	7432 .24 74122 .40 74 7433 .32 74123 .55 74 7437 .24 74125 .45 74	193 1,10 4028 .70 194 .92 4029 .90 195 .85 4030 .50
ADV26 4.74 BC157 0.06 BC2388 0.16 BD11 AF106 0.45 BC157A 0.07 BC251A 0.18 BD12 AF109R 0.36 BC158A 0.08 BC251A 0.17 BD12 AF109R 0.36 BC158A 0.08 BC251B 0.17	6 0.30 BF185 0.20 BU108 1.80 8 0.60 BF194 0.06 BU126 1.00 8 1.02 BF194A 0.07 BU133 1.75	2N1039 0.15 2N4348 2.00 2N1059 0.16 2N4448 1.50 2N1101 0.15 2N4914 3. 50	7440 .13 74128 .62 74 7441 .52 74132 .70 74 7442 .55 74141 .58 74	196 .92 4033 1.40 197 .92 4035 1.10 198 1.60 4040 .90 199 1.50 4041 .80
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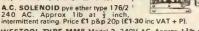


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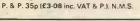
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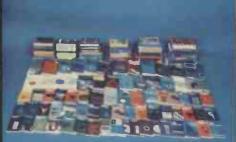
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