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Full mixing facitites with full range bass $\&$ treble controls. Alt inputs plug into standard jack sockets on from panel. Output socker on rear of chassis for an 8 ohm or 16 ohm speaker. Output in excess of 20 watts R.M.S. Very attractively finished purpose bult cabinet made from black vinyl covered steel. With a operation 200.240 volts. Size approx $12 \operatorname{lin}$. wide $x \sin ^{2}$ high $x$ 7in deep
Special introductory price $£ 28 \cdot 00+£ 2 \cdot 50$ carriage and packing.
"POLYPLANAR" WAFER-TYPE. WIDERANGE
ELECTRO-DYNAMIC SPEAKER
Size $11 \frac{1 \text { in }}{} \times 14$ hin $\times 1 \frac{\text { in deep. Weight 190z. Power handling }}{}$ $40 \mathrm{~Hz}-20 \mathrm{KHz}$. Can be mounted on ceilings ohm only. Response tables, etc, and used with or without baffle. Send S.A.E. for fül details. Only $£ 8.40$ each $+P$. \& $P$. (one 90 . two $£ 1.10$ ). Now available in either 8 in round version or $4 \frac{1}{2} \times 8 \frac{1}{2}$ in rectangular. 10
watts R.M.S. $60 \mathrm{~Hz}-20 \mathrm{kHZ} \mathrm{f5} \cdot 25+\mathrm{P}$. \& ${ }^{2}$. (one 65 p . two 75 p .).

MAINS OPERATED SOLID STATE AM/FM STEREO TUNER
 $200 / 240 \mathrm{~V}$ Mains operated Solid State F.M. A.M. Stereo
Tuner. Covering M.W. A.M. $540-1605 K \mathrm{HZ}$ V.H.F. F.M $560-1085 \mathrm{KHz}$
$8-10 \mathrm{MMz}$.
Bult-In Fernte rod aerial for M.W. Full AFC and AGC on A.M. and F.M. Stereo Beacon
Lamp Indicator. Bult in Pre. Lamp indicator. Bult in Pre control. Max o/p Voltage 600 mV R.M.S. into 20 K . Simulated Teak flnish catin will matmost any amplifier Size Biln wide 4 in high * 9 tin deep approx.
Limited number only at $£ 28 \cdot 00+£ 1 \cdot 50$ P. \& P.
10/14 WATT HI-FI AMPLIFIER KIT A stylishly finished monaural amplifier with an output of i4 watts from 2 end ech , ith push-pull. Super reproduction of both music gram allow records and announcements to follow each other. Fully shrouded section wound output transformer to match
$3-15 \Omega$ speaker and 2 independent volume controls, and separate $3-150$ speaker and 2 independent volume controls, and separate
bass and treble controls are provided giving good lift and cut. bass and treble controls are provided giving good lift and cut. instruction booklet $25 p+$ S.A.E. (Free with parts). All parts sold separately. ONLY $£ 15.50$ P. \& P, $£ 1$-40. Also available ready

STEREO MAGNETIC PRE-AMP Sens. 3 mV in for 100 mV out. 15 to 35 V neg. earth. Equ. $\pm$ idB from 20 Hz to 20 KHz . Inpu Mullard LP1159 RF-IF module 470 kHz E2.25 + P. \& P. 20p. Full specification and connection details supplied Pye VHF FM Tuner Head covering $88-108 \mathrm{MHz}$ 10.7 MHz I.F. output. $7-8 \mathrm{~V}+$ earth. Supplied pro-aligned, with full circuit diagram with precision-geared F.M. gang
A.M. Tuning gang oniy $63.15+$ P. 8 P. 35 p .

## SPECIAL OFFER

Slightly shop soiled radios by well-known manufacturer for AC Mains or battery use. MW and FM bands. Dynamic M/coll speakers, telescoplc aerial and internal ferrite aerial. Earplece
socket for personal listening. Finished in attractive simulated satherette.
Size $7^{" H} \times 9 \frac{1}{2} \mathrm{~W} \times 4^{n \prime} \mathrm{O}$ approx. Fully guaranteed

> SPECIAL OFFER
> LIMITED NUMBER ONL

GOODMANS speakers, $6 \frac{1}{\frac{1}{2}-8}$ ohm, long throw, ceramic magnet $\mathrm{E} 4-00$ each +80 p . P. \& P. (P. \& P. on two $\mathrm{f1} 120$ )

## HARVERSONIC SUPERSOUND

10 + 10 STEREO AMPLIFIER KIT
A really firsi-class Mi-Fi Stereo Amplifier Kit. Uses 14 transistors including Sithcon Transistors in the first five stages on gach channel resulting in even lower noise level with improved sensitivity. Integral Dre-amp with Bass. rreble and two vo ume simple to modily to sult magnefic cartridge-instructions included Output stage for any speakers trom 8 to 15 ohms. Compact design. all parts supplied including drilled metalwork, high quality ready
afilled printed eircuit board with component idenification cinarly drilled printed eircuit board with component idenilication cinarty marked, smart brushed anodised aluminium front panet with Simple step by step instructions enable any consiructor to build an amplifier to be proud of. Brief specification, Power output: 14 watt R.M.S. per channel into 5 ohms. Frequency response $=3 \mathrm{~dB}$ $\mathbf{1 2 - 3 0 . 0 0 0 H z}$. Sensitivity better than 80 mV Into 1 Mr . Full prower oandwidt $=$ 3os $12-160 \mathrm{~B}$ Negative feedback 18dB over inain amp. Power requirements 35 V at 1 A
Overall size 12 in wide $\times 8 \mathrm{ln}$ deep $\times 2 \mathrm{j} \mathrm{ln}$ high.
fully detalied 7 page construction manual and
klt or send 25 p plus large S.A.E. AMPLIFtER KIT POWER PACK KIT £14.50 P. \& P. 80p

CABINET
E6.00 P. 8 P. 95p
SPECIAL OFFER-only $£ 25.00$ if all 3 ltoms
ordered at one time plus E1.25P. \& P.

## HARVERSONIC STEREO 44

A solid state stereo amplifter chassis, with an output of 3-4 watts per channel into 8 ohm speakers. Using the latest high technology overload protection. All components including rectifier smoothing capacitor. fuse, tone control, volume controls, 2 pin din speaker sockets and 5 pin din tape rec./play sockel are mounted on the printed circuit panel. Size approx. 9in z 2 tin. Kiun max. depth
Supplied brand new and tested. with knobs, brushed anodised aluminlum 2 way escutcheen to allow the amplifier to be mounted horizontally or verticallyl at only $£ 10.00+50 \mathrm{p}$ P \& P. Malns transformer with an output of 17 V a.c. at 500 mA can be supplied at $\mathbf{£ 2} \cdot \mathbf{0 0}+40 \mathrm{p}$. \& P . If required. Full connection detall

## STEAEO DECODER

SIZE $2^{\prime \prime}, 3^{\prime \prime}$. In $^{\prime \prime}$ ready buith. Pre-aligned and tested for $9-16 \mathrm{~V}$ neg,
earth operation. Can be fitted to almost any FM VMF radio or tung earth operation. Can be litted to almost any FM VMF radio or tuner.
Stereo beacon light can be fitted it required. Full detalls and Stereo beacon light can be fitted it required. Fult detarls and
instructions (inclusive of hints and tips) supplieg. E600 plus 200
P. \& . Stereo beacon light if required 40 extra.

Open 9.30-5.30 Monday to Friday. 9.30-5 Saturday Closed Wednesday.
Prices and specifications correct at time of press. Subject to alteration without notice

HARVERSON SURPLUS CO. LTD.
(Dept. P.E.) 170 MERTON HIGH ST., MERTON, LONOON, S.W. 19 Tel: 01-540 3985
(Please write clearly)
PLEASE NOTE: $P$. \& P. CHARGES OUOTED APPLY TO U.K. ONLY P. AP. ON OVERSEAS ORDERS

## TADC m " TRANSFORMERS

| PRIMARY 220-240 50HZ. <br> ALTERNATIVE SECONDARY VOLTAGE AND CUPRENT AVAILABLE BY SERIES OR PARALLEL CONNECTION |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Type | Voltage | Curent | f | p/p | Type | Voltage | Current | ¢ | p/p |
| O6FEO6O8FEO612FEO620FEO650FEO66OFEO6. | ${ }^{6+6}$ | 0.54 EACH | 1.58 | 50p | 20FE24 | $24+24$ | 0.4 A EACH | 2.74 | 70p |
|  | ${ }^{6+6}$ | 0.6A EACH | 1.90 |  | 50FE24 | $24+24$ | 0.8 EACM |  |  |
|  | ${ }^{6+6}$ | 1A EACH | 2.10 | ${ }^{60}$ | 60FE24 <br> 80 FE 24 | 24+24 | 1.2 A EACH |  | ${ }^{85 p}$ |
|  | 6+6 | 1.6A EACH |  | 70 p | 80 FE24 | $24+24$ | 1.5A EACH | 4.72 | 100 p |
|  | $6+6$ $6+6$ | 3A EA | 3.25 | 70p | 60FEE 28 | $28+28$ | 0.75A EACH | 3.25 | 70p |
|  |  |  |  |  | 60FE28 BOFE28 | $28+28$ $28+28$ | 1.1A EACH <br> 1.4A EACH | 3.98 | $85 p$ <br> $100 p$ |
|  | $9+9$ | 0.5A EACH | 1.90 | $5{ }^{5}$ | 20FE 30 | $30+3$ | 0.35A EACH | 2.74 | $70^{\circ}$ |
|  | 9+9 | 0.75A EACH | 2.10 | 60 | 50 FE30 | 30+3 | 0.75A EACH |  |  |
|  | $9+9$ <br> $9+9$ | 1A EACH 2.5A EACH | 2.74 | 70 p. | 60FEE30 80 FE 30 | $30+30$ $30+30$ | 14 EACH 1.2 A EACH | 3.98 | 85p |
|  | $9+9$ | $\begin{aligned} & \text { 5A EACH } \\ & \text { 3A EACH } \end{aligned}$ | 3.98 | ${ }_{85 p}$ |  |  | 1.2A EACH |  |  |
|  |  | 0.25A |  | 50p |  |  |  |  |  |
|  | $12+$ | O.3A EACH | 1.98 | 50 c |  |  |  |  |  |
|  | ${ }_{12+}^{12+}$ | 0.5A EACH | 2.10 | ${ }^{60 p}$ | 30FE30 | O-12.120 | 1- 1 - | 3.55 | 70 p |
|  | 12+12 | 2 LA EACH | 3.25 | 70 p |  | $0-12-15$ | 2 A |  | $85 p$ |
|  | 12+ | 2.5A EACH | 3.98 | 85p | GOFE36 | 24 | 2 A |  | 85p |
|  |  |  |  | 100p | FE36 |  | 3A | 5.95 | 100p |
|  | $15+15$ $15+15$ | 0.2A EACH 0.25A EACH | 1.58 | 50p 50 p | 100FE40 | 0.12-3 $24-3$ | 4A |  | 16 |
| O8FE15 | 15+1 | O.4A EACH | 2.10 | 60p |  |  |  |  |  |
| 20FE 1550FE15 60 FE1580 FE15 |  | O.6A EACH | 2.74 | 700 |  | NTRE | AP SECO | A |  |
|  | $15+15$ $15+15$ | 1.6A EACH 2A EACH | 2. ${ }^{3.25}$ | ${ }^{70 \mathrm{p}}$ |  |  | 1 A EACH |  |  |
|  | ( | $\begin{aligned} & 2 A E A C H \\ & 3 A E A C H \end{aligned}$ | 3.98 4.72 | 85p |  | 12.0 | $1 \begin{aligned} & \text { 1A EACH } \\ & 1 / 4 E A C H\end{aligned}$ | 2.74 | 70p |
| O6FE20 |  |  |  |  |  | 15-0 | 1 A EACH | 3.25 |  |
|  | 20+20 | 0.15A EACH | . 58 | 50 p |  | 20-0-2 | 1 A EACH | 3.25 |  |
| -08FE20 | 20+20 | 0.2A EACH | 1.90 |  | 60 F 52 | 26-0-2 | 1 A EACH | 3.98 | 100 p |
|  | $20+20$ | 0.25A EACH | 2.10 | 60 D | 60FE28 | 28-0-2 | 1 A EACH | 3.98 | $100 p$ |
| 20FE20 | 20+20 | O.5A EACH | 2.74 | 70 p | 60FE60 | 30-0-3 | 1 A EACH | 3.98 | $100{ }^{10}$ |
| $\begin{aligned} & \text { GOFE2O } \\ & \text { 6OFE2O } \\ & 80 F E 20 \end{aligned}$ | 边 | 1.5A EACH |  | ${ }^{75}$ | 100 E26 | 26-0-2 | ${ }_{2 A E A C H}$ |  | 15p |
|  | $20+20$ | 2 EAC | 4.72 | - | 100 FE 36 | 36-0 | 2 A |  | 1p |
| CHARGER TRANSFORMER |  |  |  |  | R CORED AUDIO CROSS OVER COILS |  |  |  |  |
| 48FE1266FE1270 FE 212 | 0-6-12 | 4 A |  |  |  |  |  |  |  |
|  | 0-6-12 | 5A | $\left\lvert\, \begin{aligned} & 4.00 \\ & 5.010 \end{aligned}\right.$ | $85 p$ |  |  |  | 0.26 | 20 p |
|  | 0-6-12 | 6A |  |  | Fos | 0.5 mH |  | 0.30 | 20 p |
| $\begin{aligned} & \text { OBFE24 } \\ & \text { 12FE24 } \end{aligned}$ | $24+24$ | 0.15A EACH | 10 | 50 p 60 p | FE10 | 1.0 mH 2.0 mH |  | 0.60 0.75 | 25p |
|  | 4+24 | 0.2A EACH | 2.10 | 600 | FE26 |  |  |  |  |
| FLADAR ELECTRIC P.O. BOX 19 WESTCLIFF-ON-SEA ESSEX, 0702-613314 |  |  | TRADE ENOUIRIES WELCOME |  |  |  | PAYMENT TERMS: <br> C.W.O. Cheques. <br> Postal Orders |  |  |
|  |  |  | PLEASE ENQUIRE FOR OTHER TYPES NOT SHOWN |  |  |  |  |  |  |
|  |  |  |  | Postal Orders <br> Please Add 8\% VAT |  |  |  |  |  |
|  |  |  |  | After post \& $p$ |  |  |  |  |  |




## Ouality audio modules and accessories for

S450 STEREO FM TUNER Fitted with
f23.24


## FREOUENCY RANGE

SENSITIVITY BANDWIDTH
SPURIOUS REJECTION SELECTIVITY $\pm 100 \mathrm{hHz}$ AUDIO OUTPUT ( 22.5 kH 2 devialion)
STEREO SEPARATION
SUPPLY REQUIREMENTS AERIAL IMPEDANCE DIMENSIONS

88-108 MMz $30 \mu \mathrm{~V}$ 250 hH 50 dB 55 dB

The 450 Tuner provides instant programme sefection at the touch of a button ensuring accurate tuning of 4 pre-selected thations, any of which may be aftered as often as you choose simply by changing the seftings of the preset controls. stations, any of which may be altered as often as you choose, simply by changing the settin
Features include FET input stage. Vari-Cap diode luning. Switched AFC LED Stereo indicator.


| OUTPUT POWER | 7 Watts RMS |
| :--- | :--- |
| LOAD IMPEDANCE | 8 ohms |
| TOTAL HARMONIC DISTORTION | Less than $5 \%$ (Tyalcally $3 \%$ ) |
| FREOUENCY RESP ONSE | $50 \mathrm{~Hz} 1020 \mathrm{kHz} \pm 3 \mathrm{dBs}$ |
| TONE CONTROLRANGE | $\pm 12 \mathrm{dBs}$ at 100 Hz and 10 kHz |
| SENSITIVITY | 190 mV for full output |
| INPUT IMPEDANCE | 1 M ohms |
| TRANSFORMER REOUIREMENTS | $22 \mathrm{~V} . \mathrm{Cl} rated at 1 A$. |
| DIMENSIONS <br> (Less controls and panel) | $200 \mathrm{~mm} \cdot 130 \mathrm{~mm} \cdot 33 \mathrm{~mm}$ |

The Stereo 30 comprises a complete stereo pra-amplifier, power amplifiers and power supply. This, with only ine addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs b.e. high quality ceramic pich-up, stereo tuner. stereo tape deck eic. Simple to install. capable of producing really first class results.
this unit is supplied with full instructions. blach front panel, knobs, main switch, fuse and fuse holder and universal mounting brackets.


| OUTPUT POWER | 25 Watts RMS |
| :---: | :---: |
| SUPPLY | $30-50 \mathrm{~V}$ |
| LOAD IMPEDANCE | $8-16$ ohms |
| TOTAL HARMONIC DISTORTION | Less than $19 \%$ (Typically $06 \%$ |
| FREQUENCY RESPONSE | 20 Hz to 30 kHz . 2 dBs |
| SENSITIVITY | 280 mV tor full output |
| MAX. HEAT SINF. TEMPERATURE | $90^{\circ} \mathrm{C}$ |
| DIMENSIONS | $103 \mathrm{~mm} \cdot 64 \mathrm{~mm} \cdot 15 \mathrm{~mm}$ |

This high quality audio amplifier module is for use in audio equipment and slereo ampllfiers and provides output powers up to 25 RMS with distortion levels below $01 \%$


| OUTPUT POWER | 35 Watts RMS |
| :---: | :---: |
| SUPPLY | 40.60 V |
| LOADIMPEDANCE | 8-16 ohms |
| TOTAL HARMONIC DISTORTION | Lessthan 1\% (Typically $06 \%$ ) |
| FREQUENCY RESPONSE | 20 Hz 10 $30 \mathrm{hHz} \times 2 \mathrm{dBs}$ |
| SENSITIVITY | 280 mV for full output |
| MAX. HEAT SINK TEMPERATURE | $90^{\circ} \mathrm{C}$ |
| OIMENSIONS | $103 \mathrm{~mm} \cdot 64 \mathrm{~mm} \cdot 15 \mathrm{~mm}$ |

The AL80 is sumilar in design to the AL80 above and is of the same high quality but provides output powers un to 35 W with


| OPERATING VOLTAGE | $50-80 \mathrm{~V}$ |
| :--- | :--- |
| LOAOS | $4-18 \mathrm{oh}$ | 4-18 ohms $25 \mathrm{H}_{2} 20 \mathrm{kHz}$ measured

FREOUENCY RESPONSE
SENSITIVITY FOR 100 WATTS 100 Watts
$\qquad$ INPUT IMPEDANCE 33 Kohms
TOTAL HARMONIC DISTORTION 50 WATTS into 4 ohms
50 WATTS into 8 ohms
:. $06 \%$

## 

This unit. designated AL250, is a power amplifier providing an output of up to 125 W RMS, into a 4 ohm load


These low cost 10 watt modules offer the utmost In reliability and performance, whilst being compact in size.

## SPM80

 STABILISEDPOWER SUPPLY £4.40 PLY
$+35 p p=p$


| INPUT A.C. VOLTAGE | $33-40 \mathrm{~V}$ |
| :--- | :--- |
| OUTPUT D.C. VOLTAGE | 33 V nominal |
| OUTPUT CURAENT | $10 \mathrm{~mA}-1.5 \mathrm{amps}$ |
| OVERLOAO CURRENT | 1.7 amps appron. |
| DIMENSIONS | 105 mm .63 mm .30 mm |

Designed to power two AL60s at is Watts per channel 'simultaneously. Circuit Techniques include full short circult protection

PA100

\section*{| SEREEO |
| :---: |
| PRE-AMPLIFER |}


| FREQUENCY RESPONSE | $20 \mathrm{~Hz}_{2}$ to $20 \mathrm{kHz} \times 1 \mathrm{~dB}$ |
| :---: | :---: |
| TOTAL HARMONIC DISTORTION | Less than 1\% (Typically -07\%) |
| SENSITIVITY 1. TAPE <br> INPUTS 2. RADIO TUNER <br>  3. MAGNETIC P.U. | $\left.\begin{array}{l}100 \mathrm{mV} / 100 \mathrm{~K} \text { ohms } \\ 100 \mathrm{mV} / 100 \mathrm{~K} \text { ohms } \\ 3.5 \mathrm{mb} / 50 \mathrm{~K} \text { ohms }\end{array}\right\}$For an <br> output <br> 250 mV |
| EOUALISATION | Within $\pm 1 \mathrm{~dB}$ from 20 Hz to 20 hHz |
| BASS CONTROL RANGE | $\pm 15 \mathrm{dBs}$ at 75 Hz |
| TREBLE CONTROL RANGE | + 10-20 dBs al 15 hHz |
| SIGNAL/NOISE RATIO | Better than 65 dBs (All Inputs) |
| INPUT OVERLOAD | Better than 26 dBs (All inputs) |
| SUPPLY | 201040 V |
| DIMENSIONS | $300 \times 90 \times 33 \mathrm{~mm}$ (less controls) |

A top quathty stereo pre-amplifier and tone control unit, the PA 100 provides a comprehensive solution to the front end two filters for high and low frequencies

MPA30
MAGNETIC CARTRIDGE PRE.AMPLIFIER


## PRE.AMPLIFIER

The PAI2 Stereo Pre-
Amplifier chassis is desiglled allu recommended for use with the AL $20 / 30$ Audio Amplifier Modules, the PS12 Dower supply and the and Treble controls. Complete with tape output.
FREQUENCY RESPONSE $20 \mathrm{~Hz}-20 \mathrm{kHz}(-3 \mathrm{~dB})$
BASS CONTROL $\qquad$
$\qquad$

$\pm 12 \mathrm{~dB}$ at 60 Mz
NPUT IMPEDANCE
INPUT SENSITIVITY $\qquad$ $\pm 14 \mathrm{~dB}$ at $10 \mathrm{hHz}_{2}$

INPUT SENSIT
CROSSTALK $\qquad$
SIGNAL/NOISE RATIO -60 dB

OVERLOAD FACTOR
05
$\pm 20 \mathrm{~dB}$
TAPE OUTPUT IMPED ANCE
$52 \mathrm{~mm}, 84 \mathrm{~mm} \cdot 25 \mathrm{~mm}$

## PS12 POWER SUPPLY MODULE

Power supply for AL20A-30A
PA12. $\$ 450$ elc.
Transtormer T538
Input A.C. Voltage $15-20 \mathrm{~V}$. Ouipui D.C. Voltage 22-30V approx. (Dependent upon input.
Output Current 800 mA
naximum
Dimensions $60 \times 43 \times 26 \mathrm{~mm}$


## £1.90

+ 121 \% VAT + 35p p\&p.


## BP124 SIREN ALARM 5 WATTs MODULE <br> American Police screame

powered from any 12 volt supply into 4 or 8 ohm speaker. deal for car burgiar alarm, freezer break-down, and other security purposes.

## ONLY £3.50 <br> $+8 \%$ VAT $+35 p$ p $8 p$.

## MA60 HI-FI AMPLIFIER KIT

Bulld you own tod quality amplifier, save yourself Dounds. The
MA60 hit comprises the following Bl-kits modules, 2 AL60 amps, 1 ※ PA100 pre-amp, 1 A SM80 stab. Dow if supply. 1 in BMT 80 transt. oiving 15 watts RMS per channel STEREO. All modules covered by the BI-PAK sailsiaction or money bach guarantue Price $\mathrm{Ej3} 00+\mathbf{1 2 j} \%$ VAT $+62 p \mathrm{p} \AA \mathrm{p}$.

## TC60 KIT

A beautilully designed genuine TEAK WOOD veneered cabinet to put the protessional touches to your home built amplifier. Full Sockets. Noen etc Ideal for the MA60 Size: 425 mm . 290 mm . 95 mm .
Price $519 \cdot 05+12\} \%$ VAT $+15 p \rho!\rho$

## TRANSFORMERS

T538 For use with $\mathbf{S . 4 5 0}$ AL30A MPA30
Order No. 2035
 Order No. 2050 Price: $\mathbf{c 3 \cdot} \mathbf{2 5}+55 p$ pdp $+121 \%$ VAT Order No. 2034 Price: $\mathbf{5 S} 40+86 p$ pAp $+121 \%$ VAT
 Order No. 2035
2040. For use with AL60 2040. For use with AL60 Order No. 2040 2041. For use with AL80. AL120 Price $£ 5$

Order No. 2041.
Price $£ 5 \cdot 20+80 p p \& p+121 \%$ V.A.T.
CASES
TEAK $30.32 \times 23 \times 8 \mathrm{~cm}$. designed mainly for use whith our stereo 30 Audio System but has proved very helpful to home constructors. Fitted
with sotid uncut front and back. o/n $139 . \mathrm{f} \cdot \mathbf{9 5}+12 \frac{1}{2} \% \mathrm{~V}$.A.T. pRy
70 p .
TEAK $60.42 \times 29 \times 9 \mathrm{~cm}$, for use with AL60/MK60 Audio Kit. Uselul or he home constructor requiring an amplifier sleeve - has no front or back panel. o/n 140 . \& $7.00 .+12 \%$ V.A.T. P8p $85 p$.

## Professionals and Enthusiasts from BI-PAK

| 4120 50W | OUTPUT POWER | 50 Watts A.M.S. |
| :---: | :---: | :---: |
|  | SUPPIY | 70 Walls |
| avolo <br> AMPLIFIER | LOAD IMPEDANCE | 8.16 ohms |
| IWith integralheat sink and | TOTAL HARMONIC DISTORTION | 05\% Max. (Typically 02\%) |
|  | FREQUENCY RESPONSE\# 1dB | $25 \mathrm{~Hz}-20 \mathrm{kHz}$ |
| short-circuit | SENSITIVITY | 500 mV |
| f11.95 | MAX HEAT SINK TEMP. | 45 deg. C |
|  | OIMENSIONS | $192 \times 89 \times 49 \mathrm{~mm}$ |
| Introduced to fulfill the demand for a fully protected power amp. capable of driving high quality speaker systems at up to 50 w . with distortion levels below 05\%. Ideal for domestic use. Discos. P.A. systems, electronic organs etc. The generously rated com ponents ensure continuous operation at high output fevels. |  |  |
| 9M120 STABILISEO POWER SUPPLIES | AC inputs |  |
|  | SPM120/45 | $40-48 \mathrm{v}$ |
| SPM120/45 <br> SMPI20/55 | SPM120/55 | 50-L5v |
| SMP120/65 | SPM120/65 | 60-65v |
|  | OUTPUT CURRENT | 2.54 |
| $+12 \%$ V.A.T $\Rightarrow$ | RIPPLE | 1 A 100 mV 2 A 150 mV |
| SPM120 is a fixed voltage stabiliser available with an output voltage of either 45 v , 55 v , or 65 v . Designed primarity for use in audio applications. the stabiliser which provides output currents up to 2.5 A ., operates direct from a mains transformer requiring only the addition of 2 Electrolytic capacitors to complete the $\$ / \mathrm{c}$ protection. |  |  |
| GE100 Mk2. <br> 10 CHANNEL MONOGRAPHIC EQUALISER | Controt Range | $\pm 12 \mathrm{~dB}$ |
|  | Dynamic Range | 110 dB |
|  | Maximum Output | +15d8 |
|  | Frequency Hesponse | $30 \mathrm{~Hz}-20 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ) |
| £20.00 | Power Supply | 15-0-15v. |
|  | Votrage Handling Input | 3v R.M.S. |
| $\begin{aligned} & +121 \% \text { V.A.T. } \\ & \text { P. \& } 350 \end{aligned}$ | T.H.D. | 005\% |

Only $155 \mathrm{~mm} \times 65 \mathrm{~mm} \times 50 \mathrm{~mm}$ including the io $\times 10 \mathrm{~K}$ lin slider potentiometers and knubs which are mounted on a board positioned above the circuitry. In the frequency range of 31 Hz to 20 KHz you can cut and boost $\pm 12 \mathrm{~dB}$ with the 10 sliders. each o will also greatly improve the sound reproduction of your existing audio equipment. Power Supply for GE 100. o/d SG30 § 3.80 .

## VPS30

regulateo variable
STABILISED POWER SUPPLY

## $\mathbf{1 7 . 6 0 \quad \text { P. \& P. 35p }}$



## AC Input Maximum

Voltage Regulation
Aegulated Current
incorporating short circwit protection

This NEW versatile Regulated Variable Stabilised Power Supply with short circuit protection and current limiting, is a must for all electronics enthusiasts. It incorporates adjustable voltage from 2 v 30 v , with a current limiting range of $0-2 \mathrm{~A}$ with this module there is no need to build a separate power supply for each of vour projects. with the simple addition of a transtormer (o,d 2033 ). a-1ma to/d 1310 or 13051. plus a suitable shunt, a voltmetet to/d 1311 or 1306 ) a 470 ohm pot $10 / \mathrm{d} 1896$ ) a $4 \mathrm{K7} \mathrm{pot} \mathrm{to/d}$ | 1899 !. |
| :--- |
| ff's. |

## PA200

STERED
PRE-AMPLIFIER

## f16.55

$+12 \frac{1}{2} \%$ V.A.
P. $\&$ P. 40 p
The PA200 is basically our popular PA 100 . Modific

## HEADPHONES

A top quality headphone wincelyolume earpads and headband Separate balance/volume controls. Stered of Mono swith. Impedance: 8 onms. Frequency 30 . $18,000 \mathrm{~Hz} .0 / \mathrm{n} 884$. E8.70. $12 \frac{1}{\%}$ V.A.T. $\mathrm{p} \& \mathrm{p} 70 \mathrm{p}$. A brilliant compromise between price and pertorA brilliant compromise between price and pertor-
mance. Superb slereo reproduction for the mance. Superb siereo reproduction for ene
newwomer to Hi-FI. Impedance 8 onms. Frequency
$30-15.000 \mathrm{~Hz}$ o o/n $885 . \varepsilon 4.40 .+12 \frac{1}{2} \%$ V.A.T. p\&

## BIB

## HI-FI ACCESSORIES

Parallal Tracking GROOV KLEEN
The very latest in automatic record cleaning. Desig ned to suit all modern single play decks. Simple to fit it is extremely efficient. Complete with two types of
base and three height extensions. $\mathbf{6} / \mathrm{h} 8101 . \mathrm{E3.68}$ base and ihree height
Casserte Tape Eatring Kir
Casserte $T$ ape Edting Kit
Enables cassette tapes to be edited and jolned easily Enables cassette tapes to be edited and joined easily $3.2 \mathrm{~mm})$. 2 Precision Tape Cutters. Tape Pierce 9 Self-adhesive Labets. Reel of Splicing Tape. Winders and removers and instructions. all in GHOOV-STAT
GAOOV-STAT
The 818 Groov-Stat static reducer neutralises the static charge on records and other plastic surfaces. -
Cassette Head Cleaner
Essential for cleaning of tape heads, capstans and rollers. Pack contains Tape Head Applicator and tape head polisher tools. Plus bottle of special formula cleaning fluid and full
$12 \downarrow V_{0} \vee$ A.T. P\& P 35p.
 DIMENSIONS

20 Hz to $20 \mathrm{kHz} \times 1 \mathrm{~dB}$ Less than 1\% (Typicaly -70\%) 00 mv 1100 K ohms For an $3.5 \mathrm{mV} / 50 \mathrm{~K}$ ohms 500 mv ithin $\ddagger 1 \mathrm{~dB}$ from 15 CBs at 75 Hz $+10-20 \mathrm{~d} 8 \mathrm{~s}$ at 15 kHz Better than 2dBs (All inputs) $300 \times 90 \times 33 \mathrm{~mm}$ liess controis - 120

## -

Clear view edgewise meter. Centre zero
application. Sensitivity: $100 \quad 0-100 \cup 1 / \mathrm{A}$ application. Sensitivity: 100 0-100UA.
Dimensions: $45 \times 22 \times 34 \mathrm{~mm}$ 1319. E2.00. $+8^{\prime} x_{1}$ V.A.T. D\& p 35 p .

Miniature moving coil meter for stereo balance
indicator. tuning indicator for FM or similar
application. Pointer at centre indicates zero or nuil
position. Robust construction. Sensitivity:
$100-0-100 \mathrm{MA}$. Dimensions: $23 \times 22 \times 26 \mathrm{~mm}$.
/n 1318.€ $1.95+8 \%$ V.A.T. p\& p 35 p

Miniature Level Meter
Moving coil, for accurate level indication for tape recorders. amplifiers etc. Neat design. rugged construction will withstand five times rated value
Sensitivity. FSD: 200 UA. 0 . 130 UA DimenSensitivity. FSD: 200UA. Od8. 130UA. Dimen-
sions: $23 \times 22 \times 26 \mathrm{~mm}$. o/n $1320 . € 2.80 .48 \%$ V.AT. p\& 8 35p.

## umetor

Calibrated -20 so +3 and $0-100 \%$, making it suitable for use as a recording level meter or as a
power output indicator. Sensitivity: 130 uA. power outpul indicator. Sensitivity: 130 uA.
Dimensions: $40 \times 29 \mathrm{~mm}$. $/ \mathrm{n} 1321 . £ 2.00 .+8 \%$
V.A.T. p\& p 35 p .
V.A.T. P\& 35 p


## ADAPTORS

$A C$. OC enables a large range of batlery powered radios, recorder calculators to be run oft the mains. (220-240v AC) Swiichable for 6 Universal plug incorporatediof 137 E 3.60 . $+12 \% \%$ V.A.T. p\& 4 . 35 p . OC OC for use in all cars. boats elc. with pos. or neg. earth lor


## CROSSOVER NETWORKS

channels for nigh and low frequencies to correct speakers high to weeters, low to wooters. Complete with instruction
quency: $3.000 \mathrm{~Hz} .0 / \mathrm{n} 1904 . \mathrm{E} 1.10 .+12 \frac{1}{2} \% \mathrm{~V}$. p 8 p 35 p . 2.WAY tor 8 ohm speakers up 1030 watts. Frequency: 3 KHz o/n
$1905 . \mathrm{E} 1.65,+12{ }^{1} \%$ V.A.T $\$ 350$. nalls. Frequency: 800 Hz and 4.5 KHz . $\mathrm{o} / \mathrm{n} 1906$. ¢2.95. $+12 \ddagger \%$ V.A.T. p\& p . 35 p .

## MICROPHONES

or equipment requiring a high quality microphone. Sturdy. solid noulded body in black with neat chrome surround. Pick-up pattern is omnidrectional. On/Otf switch. 1 metre of tough lead with floating 25 and 3.5 mm plugs. Matching moulded strut. Impedance- 200 ohms.
Sensilvity. 90dB. Frequency: $90 \cdot 10.000 \mathrm{~Hz}$ Size: 20 mm dia $x$ $120 \mathrm{~mm} . \mathrm{o} / \mathrm{n} 1326$. £1.60 + 12 ;\% V.A.T. p\& F 35p.
DYNAMIC MICROPHONE
Superior quality portable cassette recorder mike with built-in remote witch switch and lead fitted with $5-\mathrm{pin} 240^{\circ}$ DIN plug lremote ment and 3 -pin DiN plug Imicrophonel. Provides a direct replacement tor those supplied with recorders. With detachable stand. $10.000 \mathrm{~Hz}_{2}$. Sens

RE-317: OYNAMIC MICROPHONE
Highly sensitive, high grade desk or hand mike suitable for use with many popular cassette decks. Incorporates On/Off switch and 1 meite lead with moulded standard jack plug. Complete with desk stand Ommidyectional. Impedance: 5.000 ohms. Frea response: 100 at
12.000 Hz . Sensitivity: $1-7 \mathrm{~dB}$ at $1.000 \mathrm{Hz)}$. $/ \mathrm{n} 1336 . \mathrm{E4.00}+12 \% \%$ A.T. D\& $\mathbf{D} 35$ p.

OMNIOIRECTIONAL CARDIOIO
Powered by a liv battery located within the alumnium body. Satin Silver finish with front disk protection to the diaphragm housing tem and extremely supple cable. Consumption: 0.2 mA from itv battery providing approx., 8-10,000 hours continuous life. Impedance: 600 onms Sensitivity: 70 OAB . Frequency 30.16 .000 Hz . Size: 23 mm id +267 mm , o/n 1329 . £ 12.80
UNIDIAECTIONAL CARDIOIO
Dual imp 600 and 50.000 ohms. Aesponse 50 to $14,000 \mathrm{~Hz}_{2}$.


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gooseneck chrome flexible holders
 FLOOR STAND Heavy chrome. Stow-away fee with rubber ends to,
maximum stabluty oraws to a neight of 5 'maximum. oln 1335 . c9 $50.12 \ddagger \%$ V.A.T. D8D $85 p$.
BOOM ARM for use with the above stand. Heavy chromed metal. it
gives $30^{\circ}$ reach from the stand. o/n $1337 £ 9.20 .+12 ; \%$.A.T. $\mathrm{p} \& \mathrm{p}$
Windshielo covers



## AUDIO LEADS

107
$i 118$
1
18
to pins 38,5 . Length 5.5 m . Jack connected
5 pin DiN plug to 3.5 mm . Jack connected
to pins 184 . Lengith 1.5 m .
lead. Fitted plug \& skt.
AC mains connecting lead for casselte
AC mains connecting lead for ca
recorders $\&$ radios. 2 merres
5 pin DIN phono plug to stereo
he adphone jack socket
$2+2$ pin DiN plugs to stereo jack socke
with attenuation network for stereo
headphones. Length 0.2 m
Car stereo connector. Variable geometry
plug to fit most car cassette. $\mathbf{B}$ track
cartridge \& combination units. Supplie
cartrige \& combination units. Supplied
with inline fused power lead and instructions.
60.60*


c0.85*
£0.85*
£1.10*
ع0.68*
£1.05*
$60.90^{\circ}$
60.60*

$$
\begin{aligned}
& \text { With inline fused power lead and instructic } \\
& 6.6 \mathrm{~m} \text { Coiled Guitar Lead Mono Jack Plug } \\
& \text { to Mono Jack Pluc BLACK }
\end{aligned}
$$

£1.50*
to Mono Jack Plug BLACK
3 pin DiN plug to 3 pin DIN plug. Length 1.5 m
$60.75^{\circ}$
$60.75{ }^{\circ}$
$\begin{array}{lll}5 \text { pin DIN plug to } 5 \text { pin DIN plug. Length } 1.5 \mathrm{~m} & \text { C0.75 } \\ 5 \text { pin DIN plug to Tinned open end. Length } 1.5 \mathrm{~m} & \text { Co. } 75^{\circ}\end{array}$
5 pin DiN plug to 4 Phono Plugs.

5 pin DIN plug to 5 pin OIN plug mirror
image. Length 1.5 m . 2 pin DiN plug to 2 pin DIN inline socket Length 5 m
£ $1.05^{*}$
$£ 0.68^{\circ}$
5 pin DIN plug to 3 pin DIN plug $1 \& 4$
and $3 \& 5$. Length
and 38. Length 1.5 m
2 pin DIN plug to 2 pin DIN socket. Length 10 m
co.83*
ع0.98
5 pin DHN plug to 2 phono plugs.
Connected pins 385 . Length $i .5 \mathrm{~m}$
5 pin Din plug to 2 phono sockets
5 pin Din plug to 2 phono sockets.
Connected pins $3 \& 5$. Length 23 cm
5 pin DIN socket to 2 phono plugs.
Connected pins $3 \& 5$. Length 23 cm
Coiled stereo headphone extension lead. Black. Length 6 m
AC mains lead for calculators etc.
c0.75*
c0.68* AC mains lead for calculators etc.
C1.75*

T. p8p 35p. ofn 133

DEPT. PE5, P.O. Box 6, Ware, Herts. Components Shop: 18 Baldock Street. Ware, Herts.

## KITS FOR SYNTHESISERS, SOUND EFFECTS <br> COMPONENTS SETS include all necessary resistors. capacitors. semiconductors, potentiometers and ransformers. Hardware such as cases, sockets, knobs, be bought separately. Fuller details of kits. PCBs and parts are shown in our lists. <br> PHONOSONICS


P.E. MINISONIC MK. 2 SYNTHESISER

A portable mains-operated Miniature Sound Synthesiser. with keyboard circuits. Although having slightly fewer facilities than the large P.E. Synthesiser the functions offered by this design give it great scope and versatilty. Consists of $2 \log$ VCOs. VCF. 2 envelope shapers, 2 voltage oscillator and detector, ring modulator, noise generator mixer, power supply.
Set of basic component kits (excl. KBD R's
and tuning pots - see list for options availabiel. from $\mathbf{£ 6 1 . 0 0}$
P.E. SYNTHESISER (P.E. Feb. 73 to Feb. 74)

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Enables a voltage controlled synthesiser to automatically play pre-programmed tunes of up to 32 pitches and 128 notes long. Programs are keyboard initiated and note length and rhythmic pattern are externally variable. (Please use order codes quoted in brackets.)

$$
\begin{array}{lr}
\text { Main Circuit (Nov) excl. sws (KIT 76-1) } & \text { £ } 18.03 \\
\text { Power Supply (KIT 76-3) } & \text { £4.72 }
\end{array}
$$

Power Supply (KIT 76-3)

Trigger Inverter and Alt. Output (KIT 76-2)
LED Counter (KIT 76-4)
PCB (as perlished) for KI
PCB (as published) for KITS 76-1 \& 3 (PCB 76A)
PCB for KITS 76-2 \&
PCB for KITS 76-2 \& 4 (PCB 76B)

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Tone Generator (KIT 77-2)
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Diode Gates (KIT 77-3)
Chorus Generator (KIT 77-4)
Voicing System (KIT 77-5)
C8.77
£ 14.66
$£ 18.81$

Voicing System (KIT 77-5)
Double Pider
Double-sided PCB for Power Supply, Tone Generator \& Diade Gates with most of the Matrix wfring as printed tracking (PCB 77UR) $\quad £ 18.40$ PCB for Voicing System (PCB 770 )
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$£ 18.40$
$£ 2.65$

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The basic five octave electronic piano (P.E. May/Sept 75 and Sound Design) has switchable alternative voicings for HonkyTonk. ordinary piano, and Harpsichord or a mixture of any of these three, together with facilities including fast and slow tremolo loud and soft pedal switching, and sustain pedal switching. The modification retains all the circuitry associated with the piano but in addition provides an organ-voice envelope facility with 5 switchable pitches, variable attack and sustain. phasing and vibrato.
Set of components (excl switches) for PSU. Frequency generator, Pitch and Note Divider, Envelope Shapers, Voicings and Control circuitries. (Order as KIT 71-5) $\quad$ ) $99-25$
Sel of PCBs lorder as PCB SET 71-6
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Modulates the attack. decay and filter characteristics of an audio signal not only from a guilar but from any audio source. producing 8 different switchable effects that can be further modified by manual controls. Possibly the mos interesting of all the low-priced sound effects units in our
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Component set with special foot operated switches $\quad \mathbf{~} 7.69$ Alternative component set with panel switches Printed circuit board

CIRCUIT AND LAYOUT OIAGRAMS are supplied Tree with all PCBs unless "as published"
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## ELEKTOR ELECTRONIC PIANO (Elektor Sept 78

A touch-sensitive, multiple-voicing 5 octave piano using the
1atest integrated-clrcuit techniaues for latest integrated-circuit techniques for the keving and
envelope shaping and virually eliminating bee-hive, noise hitherto inherent in previous electronic planos. Oetails in our lists.
digital reverberation Unit (Elektor May 78)
A very advanced unit using sophisticated i.c. techniques instead of mechanical spring-lines. The basic delay range of
24 to 90 mS can be extended up to 450 mS using the extension unit. Further delays can be obtained using more extensions.

> Main component set (KIT 78-1 Kxtension component set (K1T $78-2$ PCB for Kit 78.1 (PBE 78A) PC8 for Kit $78-2$ (PCB 78B)

| E45.45 |
| :--- |
| F 43.36 |

$f 43.36$
$£ 2.86$
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Main component set (KIT 83-1)
Addtional Delay Set (KIT 83-2)
$\mathbf{f} 26.18$
$\mathbf{f} 18.25$
PCB (as published) to hold both above
RESONANCE FILTER (Elektor Oct 78)
This filter module has been designed to allow a synthesiser to produce a more realistic simulation of natural musical
Basic component set (KIT 82-1)
¢15.10
f3.29
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(P.E. Oct 78)

This unit allows external inputs, such as guitars, microphones etc. to be processed by the circuits within a
Basic component set (incl PCB) (KIT 81-1) $\quad \mathbf{£ 2 . 9 4}$
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An extremely versatile sound processing unit capable of producing, for example, Flanging, Vibrato. Reverb, Fuzz and
Tremolo as well as other fascinating sounds. May be used with most electronic instruments. Details in our lists.
RHYTHM GENERATOR KITS
Several available - details in our lists.
GUITAR FREQUENCY DOUBLER (P.E. Aug. 77)
modified and extended version of the circuit published.
Compnnent set and PCB

## GUITAR SUSTAIN (P.E. Oct 77)

Maintalns the natural attack whilst extending note duration.
Component set, PCB and foot switches
Component set, PCB and panel switches

## WIND AND RAIN UNIT

A manually controlled unit for producing the above-named sounds.
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Sophisticated. versatile Fuzz unit. including varlable and switchable controls affecting the fuzz quality whist retaining the attack and decay, and also providing filtering. Does not be used with it and with other electronic instruments Component set using dual slider pot Component set using dual rotary pot Printed circuit board

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Simple Fuzz unit based upon P.E. "Sound Design" circuit.

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Gives a much shriller quality to audio signals fed through it
The depth of boost is manually adjustable.
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Component set (incl. PCS) (Order as Kit 60-1) $\quad$ E3.64
ENVELOPE SHAPER WITHOUT VCA (P.E. Oct. 75)
Provides full manual control over attack. decay. sustain and release functions. and is for use with an existing voltage ontrolled ampltfier.
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ENVELOPE SHAPER WITH VCA (P.E. Apr. 76) This unlt has its own voltage controlled amplifier and has full
manual control over attack. decay. sustain and release funcrions

Component set (incl. PCB)
TRANSIENT GENERATOR (P.E. APr. 77)
An envelope shaper, without VCA, having the usual attack decay. sustain and release functions. and in addition it also provides a Hepeat Effect enabling a synthesiser to be programmed to Imitate such instruments as a mandolin or banjo.
Component set

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Alightly modified version of the circuit published in
Elektor automatic control over the rate of phasing and vibrato. Component set
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$\mathbf{~} \mathbf{2} .33$
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A simple but effective manually controlled unit for muslc mound into live or recorded Component set (incl. PCE)

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Type GA: 1 pair of contacts, normally open
Type GB: 2 pairs of contacts, each pair normally open
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Type 4PS: 3 pairs of contacts plus single-pole changeover

Printed Circuit Boards for use with most contacts (thus eliminating much interwirngl are available. Details in our lists.
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P.E. TUNING FORK (P.E. Nov. 75)
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Main component set lincl. PC8)
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omponent set (incl. PCB)

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Details in lis1s.

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\hline 7400 & 13p & 7413 & 22p & 7454 & 14p & 7485 & \(74 p\)
\(27 p\) & 74121 & 27p \\
\hline 7401 & 13p & 7414 & 60p & 7460 & \(14 p\) & 7490 & 40 p & 74141 & 54 p \\
\hline 7402 & 13p & 7420 & 14p & 7470 & 24p & 7491 & 71p & 74151 & 60p \\
\hline 7403 & 13p & 7430 & 14p & 7472 & 24p & 7492 & \(46 p\) & 74154 & 1.60 \\
\hline 7404 & 18p & 7440 & 14p & 7473 & 23p & 7493 & 40p & 74190 & 940 \\
\hline 7405 & 14p & 7442 & 54p & 7474 & 23p & 7494 & 66p & 74191 & 94p \\
\hline 7407 & 22p & 7443 & 60p & 7475 & 45p & 7495 & 57p & 74192 & 94p \\
\hline 7408 & 18p & 7444 & 60p & 7476 & 32p & 7496 & 63p & 74193 & 94p \\
\hline 7409 & 18p & 7447 & 70p & 7480 & 41p & 74100 & 73p & & \\
\hline 7410 & 14 p & 7450 & 14p & 7482 & 61p & 74104 & 40p & & \\
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\hline \multirow[t]{2}{*}{} \\
\hline \\
\hline
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\begin{tabular}{l} 
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\hline TTL (see catalogue for full range)
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Flat Nose Pliers Flat Nose Pliers...
Snipe Nose Pllers End/Top Cutte
Also in slock
BAHCO Quality Toot range. \begin{tabular}{llll}
\multicolumn{5}{l}{} \\
LINEAR /see Catalogue for full rangel \\
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\end{tabular}

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\section*{LUCRATIVE MARKET}

FROM time to time we see large industrial distributors or manufacturers breaking into the component retail market; this has happened again just recently. They obviously feel the market is a lucrative one; they certainly enter it with both eyes open and usually after considerable researchsometimes with quite a fanfare too! Unfortunately, nearly all these companies have in the past come to grief, having struggled along for a few years making little or no profit and finally giving up and closing down.

We were fortunate enough to be invited to the press launch for Doram-it was an impressive affair in a wine cellar in the city of London. The link between the fledgling company and Radiospares-as RS Components were then known-was emphasised. There was a display of kits and components that Doram would market and directors and technical staff were on hand to answer the many questions from the hobbyist and general electronic press. Those who doubted the viability of the project, in the light of the failure of other companies, were soon quietened by the enthusiasm and professionalism of the Doram people.

Alas it was not to be, and having struggled along for a few years, a recent change in policy to get out of components and just sell kits was eventually squashed by the decision to shut up shop completely. How then can so many "private" firms continue.

Perhaps it is simply that "industrial" companies need a bigger profit margin? Maybe they use over complex systems, designed for industry and not mail order? Or could it simply be that their marketing is all wrong for the hobbyist? It's probably a combination of all these factors. One thing is clear, our market is not an easy one to judge, or to be profitable in. Of course, we wish all "new" companies well and we hope the latest venture by a distributor 'will be successful-only time will tell.

\section*{FRONT ROOM SUPPLIERS}

Many successful cómponent suppliers advertising in this magazine have started with one or two people, often from someone's front room or a small back street office. They have built up a business from a spare time occupation and now many of them employ upwards of a dozen people. Perhaps because they started with very little capital they have learnt how to keep the overheads down, how to market to
you the hobbyist consumer, and above all how to work and make a profit on a low margin.

We do, of course, get some complaints about suppliers, long delivery times, etc. but they are often faced with no delivery from manufacturers and the breaking of promised delivery dates from the world's suppliers. There is the odd bad company, but rest assured they are quickly weeded out of our pages and our mail order protection scheme is used to help those who's money has gone to a bankrupt company.

By and large we have nothing but praise and admiration for the component suppliers; most of them operate a fast, efficient service at an excellent price. They carry many thousands of different items with a large proportion available for pence rather than \(£ \mathbf{s}\). It is 'necessary for them to buy stock directly from many countries, to rely on the post for most of their sales and to operate in a highly competitive field. Industry has shown many times that it is willing, but often unable to operate successfully in this area, so without the private retail component suppliers our hobby would probably cease, certainly be more expensive.

Mike Kenward

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\title{
Microprocessor Evaluation System \\ D.S.COUTTS \\ (General Instrument Microelectronics)
}


\section*{Part 1}

WITH microprocessors steadily spreading through the electronics hobby scene, the author felt that it was time to learn a little about them and how they can be programmed to perform simple tasks. The question was, where to start? The requirement was for a very simple machine using a powerful processor so that the lessons learned could be applied later to build a more sophisticated machine. The simplest approach seemed to be to use switches to read in the address and data information (eliminating the need for a PROM programmer) and to use single l.e.d.s as the output display. The microprocessor chosen was the GENERAL INSTRUMENT CP1610, a 16 bit machine with a very powerful instruction set. A large memory wasn't considered necessary to learn the rudiments


Fig. 1.1. Block diagram of the Evaluation System
of programming so four 2112 RAMs were used, giving a total of 256 sixteen bit words ( 1,024 bits).
The basic system block diagram is shown in Fig. 1.1. It consists of a power supply, the CP 1610 CPU, TTL clock, reset and control circuits, four 2112 RAMs and TTL address and output latches. Data and addresses are set up by 16 s.p.d.t. toggle switches each with a centre off position. There is also a LOAD/RUN toggle switch and two push buttons for ADDRESS ENTRY and DATA ENTRY. 16 single l.e.d.s are used for the output. The memory occupies address space 8-377 OCTAL ( ( - 255 BINARY) and the output latches are at address 377 OCTAL ( 255 BINARY).

\section*{INTRODUCTION TO CP1610}

The CP1610 internal block diagram is shown in Fig. 1.2. There is an array of 8 sixteen bit registers. Register 6 serves as the stack pointer (SP) while register 7 serves as the program counter (PC). The other registers are general purpose and may be used for arithmetic or other operations.

Communication to the outside world is via a 16 bit bidirectional bus, but internally the data is processed in two 8 bit bytes. There are four status flags SIGN (S), ZERO (Z), OVERFLOW (OV) and CARRY (C), and there is a 10 bit instruction register. Table 1 shows the CPU pin connections and functions.

All the pins are not used in our basic processor, pins 27 , \(28,31,32,33\) are linked to +5 V and pin 40 also goes via a resistor to +5 V . Pins \(1,22,23,24,25\) are not used at first. This leaves us with four power input pins, two clock input pins, a reset input, sixteen I/O lines for data and address and three output pins to decode for bus control purposes. Not too frightening \(\mid\)


Fig. 1.2. Internal block diagram of the CP1610
The processor instruction set is shown in Table 2. It may seem a bit overpowering at first but there is a glossary of terms at the end of the table and it is intended to give programming examples when the processor has been completed.

\section*{CIRCUIT DESCRIPTION}

The complete circuit diagram of the system is shown in Fig. 1.3. The circuit is relatively straightforward and will be described in blocks.

\section*{CLOCK CIRCUIT}

IC1 \(a, b\) and \(c\) are used for the master oscillator and this feeds the shift register \(2 \mathrm{a}, \mathrm{b}\) and \(3 \mathrm{a}, \mathrm{b}\). This shift register is used to produce four time slots, TS 1 to TS4. IC5 \(a\) and \(b\) are used to produce two non-overlapping clocks \(\varnothing 1\) and \(\varnothing 2\). These clocks are buffered by the open collector TTL IC6a and \(b\), and drive the CPU via emitter followers TR 1 and TR2.

\section*{LOAD/RUN AND RESET CIRCUIT}

When S1a is in the load position IC4 pin 8 is low, preventing the shift register from running and stopping the clocks to the processor. The low on IC4 pin 8 is fed to IC4 pin 5 , holding pins 6 and 2 high. As pin 1 is also high, pin 3 is held low, holding the CPU in the reset state. In this state the sixteen processor I/O lines are in their high impedance state and the switches ( \(\mathbf{S 4}\) to S19) may be used to load the program into RAM. When S1a is changed to the run position the CPU clocks are enabled and the CPU reset line goes high when IC3a pin 9 goes low, and the program is executed.

\section*{BUS CONTROL}

IC8, a B.C.D. (decimal decoder) is used to decode the \(B C 1, B C 2\) and BDIR lines from the processor to produce the necessary bus control signals. BC1 is on pin 3, BC2 on 4 and BDIR on 5. Pin 12 is fed from IC5c and goes high during reset and also during time slot TS1. This prevents IC8 from outputting unwanted signals during reset and also during

\section*{VCC \(\quad+5 V\) \\ VBB \(-3 V\) \\ VDD \({ }^{\circ}+11 \mathrm{~V}\) \\ D0-D15 DATA AND ADDRESS LINES \\ MSYNC RESET LINE}

EBCAO EXTERNAL BRANCH CONDITION
EBCA1 ADDRESS. Outputs of 4 least significant
EBCA2 bits of instruction register, used externally
EBCA3 to select 1 of 16 digital states to be sampled by the CPU during BRANCH EXT. instruction.

EBCI EXTERNAL BRANCH CONDITION INPUT

\section*{BDRDY}

Input, used to make CPU wait for slow memories.
\(\left.\begin{array}{l}\text { INTR } \\ \text { INTRM }\end{array}\right\}\) Interrupt inputs, one maskable
TCI Output, terminates current interrupt.
BUSRQ Used to cause CPU to relinquish control of bus.
BUSAK Output to signal that CPU has released bus.
STPST Alternately stops and starts CPU.
HALT Output, indicating CPU halted.
PCIT 1. As an input prevents the program counter incrementing during fetch phase of instructions.
2. As output, outputs negative pulse during SIN (software interrupt) instructions.
\(\left.\begin{array}{ll}\text { BC1 } & \text { Bus Control 1 } \\
\text { BC2 } & \text { Bus Control } 2 \\
\text { BDIR } & \text { Bus DIRection }\end{array}\right\}\)\begin{tabular}{l} 
see bus control \\
signals below
\end{tabular}


\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|r|}{\multirow[b]{2}{*}{}} \\
\hline & \\
\hline & \begin{tabular}{l}
 \\
 \\

\end{tabular} \\
\hline & \(=\sim=\) \\
\hline & － \\
\hline & \\
\hline
\end{tabular}
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{8}{|l|}{REGISTER－REGISTER} \\
\hline mnemonic & Optrand & craces & \multicolumn{3}{|l|}{Instruction} & DESCCAPTION & STATUS
CHANGE \\
\hline move & sss． 000 & \(6^{\circ}\) & 0010 & sss & 000 & MOVI contents of Register SSS to register DDD．＂I DDD is & 5． 2 \\
\hline \({ }^{\text {TSTR }}\) & sss & \(6^{\circ}\) & 0010 & sss & sss & TeST contents of Register SSS．＊If SSS is \(\mathbf{B}\) or 7 add to Cycles & 5．2 \\
\hline JR & sss & 7 & 0010 & sss & 111 & Jump to address in Register SSS．\｛Move address to & 5.2 \\
\hline A00 & sss． 000 & 6 & 0011 & sss & 000 & ADD contents of Register SSS to contents of register DOD． & 5．2．c．ov \\
\hline SUEP & sss． 000 & 8 & 0100 & sss & 000 & SUBtract of Register SSS from contents of register DOD． Results to DOD & s．2． C ． ov \\
\hline CMPR & sss． 000 & \({ }^{8}\) & 0101 & sss & 000 & CoMPare Regisise SSS with register 000 by subtraction & s．z． C ．ov \\
\hline amor & Sss． 000 & ¢ & 0110 & sss & 000 & logical AND contents of Register SSS with contents of register DOD．Results to DOD & s． 2 \\
\hline хоaR & sss． 000 & © & 011 & sss & 000 & eXclusive OR contents of Register SSS with contents of register 000．Results to 000. & 5． 2 \\
\hline Clars & 000 & \({ }^{6}\) & 0111 & 000 & 000 & Cleare Registe 10 ero． & s． 2 \\
\hline  & 000
000 & \({ }_{6}^{6}\) & 0000
0000 & 001
0010 & 000
000 &  & 5．\({ }_{\text {s．}}\) \\
\hline comb & 000 & 6 & 0000 & 011 & 000 & Oneis COM piement contenis of Reqistec 000 ．Resulis to 000. & \\
\hline NEGR & 000 & 6 & 0000 & 100 & 000 &  & s．2．c，ov \\
\hline AcCR & 000 & 8 & 0000 & 101 & 000 & AOS Cary bit ts contents of Reyister OOD．Results to 000． & s．z．c．ov \\
\hline
\end{tabular}
\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline \multicolumn{8}{|l|}{\begin{tabular}{l}
REGISTER SHIFT Executable only with Register 0，1，2，3．n＝1 or \(\mathbf{2}\)（places） \\
Shift Right instructione set the \(\mathbf{S}\) flip－flop with Bit 7 of the result after the Instruction． \\
Add 2 cycles Hf shift is 2 bltg or two bytes． \\
Shifts are not interruptable．
\end{tabular}} \\
\hline SWAP &  & 8 & 0001 & 000 & Nคค & \begin{tabular}{l}
\(N=0\) ．SWAP brtes of register RR．\(S\) equals Bin 7 of results of SWAP． \\
\(N=1\) ，SWAP brtes of registet RR，then swep them back to orginal lorm
\end{tabular} & s， 2
s． 2 \\
\hline SU & R月＜，n＞ & \({ }_{8}^{6}\) & 0001 & 001 & NRR & \begin{tabular}{l}
\(\mathrm{N}=0\) ，Shith Logical Leth one bit，zero to low bit． \\
\(N=1\) ．Shith Logical Left two bits，zero to low 2 bits
\end{tabular} & \[
\begin{aligned}
& 5.2 \\
& 5.2
\end{aligned}
\] \\
\hline RLC & RRイ，n＞ & 8 & 0001 & 010 & N月R & \begin{tabular}{l}
\(\mathrm{N}=0\) ，Rotate Leth one bit using Carry bit as bit 18. \\
\(\mathrm{N}=1\) ．月otate Left two bits using C as tit 17 and OV as bit 16
\end{tabular} &  \\
\hline Suc & RRイ．n＞ & \({ }^{6}\) & 0001 & 011 & NRR & \begin{tabular}{l}
\(\mathrm{N}=\mathbf{0}\) ．Shith Logical teh one bit using C as bit 16 ． zero to low bit． \\
\(\mathrm{N}=1\) ，Shift Logical teft two bits using C as bit 17 ． OV as bit 16. zelo 10 low 2 bits
\end{tabular} & s．2．C
s． 2 C .0 Ov \\
\hline SLR & RR＜n＞ & 8 & 0001 & 100 & NRR & \(\mathrm{N}=0\) ．Shith Logital Right one bit，zero to high bit \(N=1\) ．Shih Logical Right two bits，zero to high two bits． &  \\
\hline SAA & RR＜， \(\mathrm{n}^{\text {＞}}\) & 8 & 0001 & 101 & NRR & \(\mathrm{N}=\mathbf{0}\) ．Shitt Arthmetic Right one bit，sign bit copied to high bit． \(\mathrm{N}=1\) ．Shift Arithmetic Right two bits，sign tit copied to high bits & 5.2
5.2 \\
\hline RRC & RRC．a＞ & & 0001 & 110 & NRR & \begin{tabular}{l}
\(\mathrm{N}=0\) ．Fotate Right ont bit using Carry as bit 16. \\
\(N=1\) ．Fotate Right two bits using \(C\) as bit 16 ． OV as bit 17 ．
\end{tabular} & \[
\begin{aligned}
& \text { s.Z.C } \\
& \text { S.Z. C. } \mathrm{OV}
\end{aligned}
\] \\
\hline SARC & RA＜，\({ }^{\text {P }}\) & 8 & 0001 & 111 & NRR & \begin{tabular}{l}
\(N=0\) ，Shitf Aithmetic Right one bit．thru Cerry．sign tit copied to thigh bit \\
\(N=1\) ．Shith Anthmetic Right twe bits，thru Cary and OV． sign bit copied 10 high 2 bhs
\end{tabular} & \[
\begin{aligned}
& \text { s. L, C } \\
& \text { s.z.c.ov }
\end{aligned}
\] \\
\hline
\end{tabular}
\begin{tabular}{|c|c|}
\hline &  \\
\hline &  \\
\hline \multirow[b]{3}{*}{} &  \\
\hline & 8 \(\hat{\hat{V}}\) \\
\hline &  \\
\hline
\end{tabular}








COMPONENTS . . .

\section*{Resistors}
\begin{tabular}{ll} 
R1, R2, R14, R15, R16 & 1k (5 off) \\
R3-R13, R17, R50 & 10k (13 off) \\
R18-R33 & 100 (16 off) \\
R34-R49 & 330 (16 off) \\
R51-R54 & 390 (4 off) \\
R55 & 3 k 9 \\
R56 & 1 k 8
\end{tabular}

\section*{Capacitors} C1
C2, C3
C4-C17
C18
C19, C20
C21
C22
C23
C24

\section*{Diodes}

D1-D16
D17. D18
D19, D20
D21, D22, D23
D24
D25

Transistors
TR1, TR2
TR3

\section*{Integrated Circuits}
\begin{tabular}{ll} 
IC1, IC27 & \(7404(2\) off \()\) \\
IC2. IC3 & \(7474(2\) off \()\) \\
IC4, IC7, IC10 & \(7400(3\) off \()\)
\end{tabular}
\begin{tabular}{ll} 
IC5, IC9 & 7410 (2 off) \\
IC6 & 7406 \\
IC8 & 74145 \\
IC11 & 7402 \\
IC12 & 74 L30 \\
IC13, IC14 & 74175 (2 off) \\
IC19, IC20, IC21, IC22 & 74 L75 (4 off) \\
IC15, IC16, IC17, IC18 & 2112 (4 off) \\
IC23 & CP1610 \\
IC24, IC25 & 7805 (2 off) \\
IC26 & 7403
\end{tabular}

\section*{Switches}

S1
S2, S3
S4, S19
S20
S21-S24
S25

3 pole change over (toggle)
S.p.c.o. push button (2 off)
S.p.c.o. centre off (16 off)
S.p.s.t. 1 A

Push button ( 4 off)
S.p.d.t.

Miscellaneous
FS1 \(\quad\) 1A slow blow fuse 20 mm
T1 \(\quad 240 \mathrm{~V}, 6 \mathrm{~V} 1 \mathrm{~A}, 6 \mathrm{~V} 1 \mathrm{~A}\)
XL1 \(\quad 1 \mathrm{MHz}\) crystal
Case \(\quad(380 \times 200 \times 100 \mathrm{~mm})\)
Case \(\quad(150 \times 80 \times 40 \mathrm{~mm})\)
Fuse Holder
Holders for i.c.s. (if req.)
Printed circuit board (main board)
Printed circuit board (power supply)
LS 1 8ohm 0.2 W speaker
7 way DIN socket (2 off)
7 way DIN plug (2 off)
Veroboard

\section*{CONSTRUCTORS NOTE}

All the components including the CP1610 microprocessor are available together with a data sheet from Technomatic and Watford Electronics who both advertise in the magazine.

TS1 when data is not valid. IC8 pin 2 outputs ADAR which causes the addressed contents of memory to be gated on to the bus and strobed into the ADDRESS REGISTER. Pin 5 outputs BAR which strobes the CPU output into the ADDRESS REGISTER. These two low going signals are orred together by IC9a, and used to strobe the address into IC13 and 14, the address latches. IC8 pin 3 outputs IAB which, via gates \(11 \mathbf{a}\) and \(\mathbf{b}\), enables the RAMs IC15 to 18 , to output the program starting address on to the bus. IC8 pin 4 outputs DTB. This signal is fed via gates \(9 b\) and 11 b to the RAMs chip enable to read data out on to the bus. The final output of IC8, pin 7, is DWS. This signal also enables the RAMs via gates 9 b and 11 b , but it also strobes the RAMs write inputs via gates 9 c and 10 a . It is also used to strobe data into the output latches IC19 to 22 when gated with address 255 in gate 11 c .

\section*{ADDRESS AND DATA ENTRY}

When S1 is in the load position S 1 b provides +5 V to one bus feeding the switches S4 to S19 while S1c provides OV to the other bus. Switches S4 to S19 are normally at their centre off position. To load an address into the address latches, first set the switches to the required binary pattern \(\left.1+5 \mathrm{~V}={ }^{\prime \prime} 1^{\prime \prime}, 0 \mathrm{~V}={ }^{\prime \prime} 0^{\prime \prime}\right)\) and then push the address enter button S2. S2 feeds the anti bounce circuit IC7a and.b. IC7 pin 8
goes low; this puts a low on IC9 pin 1. IC 10 pin 6 goes low, reading the data from the switches into the address latches when it goes high again.

The word that has been loaded into the address latches is now addressing the RAMs although they are not yet enabled. By setting the sixteen switches to the required DATA pattern and pushing the ENTER DATA switch S3, the data is entered into the addressed RAM location. When S3 is pushed IC7 pin 6 goes low and this enables the RAMs via gates \(9 b\) and 11 b . At the same time it strobes the RAMs write enable via gates 9 c and 10a, writing the data from the sixteen switches into the RAMs.

\section*{OUTPUT LATCHES}

The output latches (IC19-IC22) are. 74L75 quad latches. The 16 bus lines feed the D outputs of the latches. When binary address 255 is selected (OCTAL 377) IC12 pin 8 goes low and if DWS goes low at the same time the output of gate 11 c goes high, strobing the data from the bus into the latches. The output latches provide both Q (high going) and \(\bar{X}\) (low going) outputs. The \(\overline{\mathrm{Q}}\) outputs are used to drive single l.e.d.s via current limiting resistors. The \(\mathbf{Q}\) outputs of the latches could be brought out to drive external circuitry if required.
NEXT MONTH: Construction details

\title{
VFHICHE INSIMRUMI DNII DISPMAYS
}

\begin{abstract}
The car market is acknowledged to be one of the most difficult for electronic systems to penetrate. The combined environment of temperature, humidity, vibration, electrical transients, etc. coupled with the required visual appeal and cost targets, are very demanding. Nevertheless, the present state of the available technologies indicate that displays will be available for vehicles in the early 1980s.
\end{abstract}

THE modern motor vehicle contains instruments that have been optimised to provide good performance at reasonable cost levels. They are usually attractively mounted in plastics mouldings and form an integral part of the vehicle fascia. The content of the instrument package is defined by the vehicle specification, written around the marketing requirements, on the generalised assumption that the cost is directly related to the complexity.

Electronic instruments in the form of tachometers and speedometers, etc. have been introduced in the past few years which utilise medium-scale integrated circuits, but again the display system used is a moving-coil instrument that is essentially of mechanical construction. The manufacturing cost of an individual instrument is fairly low, and to consider utilising electro-optic methods of display on an individual basis would generally be too expensive.

\section*{BASIC REQUIREMENTS}

Examination of available display technologies indicates that separate instruments, complete with drive electronics, would be more expensive than their conventional electromechanical equivalent. If, however, all the individual elements of a dashboard are combined on one large display, and the electronics are also integrated to form one complete panel, then a cost-effective solution can be realised. The phrase "solid-state displays" has been coined to cover this type of panel, though it can be a misnomer in certain cases.

The most important parameters to be considered when choosing a particular technology are those of overall system cost, visual appeal, reliability, life and legibility, which covers the brightness and contrast aspects. The display must be capable of
being manufactured on a large scale with a reasonable level of tooling and a minimum of assembly labour. Furthermore, the display technology chosen must be compatible with available integrated circuits and operate within reasonable voltage and power limitations. The ability to multiplex becomes a very critical factor.
The main parameters of importance for choosing a display technology suitable for a motor vehicle are those of total system cost, visual appeal, reliability, including operation over a wide temperature, vibration, and humidity range, together with a life requirement equal to that of the vehicle.

\section*{DISPLAY TECHNOLOGIES}

Generally, the available technologies break down into those suitable for small-area display and those suitable for large areas; they also have different suitability as regards the optimum number of bits. A "bit" of information is defined as a single separate display area that has to be controlled independently (Table 1).

\section*{DISPLAY COST COMPARISON}

The graphs in Fig. 1 show how the cost of these displays differs depending on their manufacture. A display consisting of only a few bits, e.g. warning lights, will have the lowest cost using individual elements like l.e.d.s. A display with an extremely large number of bits will have the lowest cost using a cathoderay tube.

\footnotetext{
* Brian Shepherd is the Engineering Director of the Vehicle Instrumentation Division, Smiths Industries Ltd., Cricklewood Works, London NW2 6NN
}

\section*{TABLE 1. OPTIMUM NUMBER OF BITS FOR A} GIVEN DISPLAY TECHNOLOGY
\begin{tabular}{lr}
\hline Technology & \begin{tabular}{r} 
optimum \\
number of bits
\end{tabular} \\
\hline Tungsten filaments & \(1-20\) \\
Light-emitting diodes & \(1-30\) \\
Cathode-ray tubes & \(10 k-250 k\) \\
Gas discharge (plasma panels) & \(30-5 k\) \\
A.C. or d.c. electroluminescence & \(30-3 k\) \\
Liquid-crystal display & \(5-200\) \\
Electrochromic (liquid or solid) & \(5-200\) \\
Electrophoretic & \(5-200\) \\
Vacuum fluorescent & \(10-100\) \\
\hline
\end{tabular}

A typical full instrument panel for a vehicle will have between 150 bits for a low-line vehicle to 300 bits for a high-line vehicle. It is evident that d.c. electroluminescence (d.c.e.l.) is most cost effective within this range, and that other technologies, e.g. liquid-crystal display (l.c.d.) and vacuum fluorescence, could also prove viable. The system of course also includes the drive electronics, and thus the system cost will involve other technical considerations as described later.

\section*{D.C. ELECTROLUMINESCENCE}

The particular advantage of d.c.e.l. for vehicle-display applications is that, under normal and low ambient lighting, the display has considerable visible appeal with a reasonably wide range of colours. The disadvantages at present are associated with the available brightness and contrast ratio under extreme sunlight conditions. Work is continuing to overcome these problems.

The technology emits light and hence requires no other means of illumination. The construction is outlined in Fig. 2. The overall visual appeal, together with package cost advantages, have made it the favourite technology at present.
D.C. electroluminescent displays are based on polycrystalline copper and manganese-doped zinc-sulphide power phosphors. The devices are fabricated on transparent conductively coated glass substrates, which are photolithographically etched to provide delineation of the required light emission pattern and electrical input. All displays are hermetically sealed in a robust flat-pack configuration.

As a display technology, d.c.e.l. possesses many unique and advantageous features. The devices operate with high visual efficiency giving excellent legibility and an arresting appearance. The emission spectrum is broad band with its peak close to that of the human eye sensitivity curve. The basic bright yellow colour may be externally filtered to provide readily discriminated green and red displays. The emission is physiologically restful for continuous observation and contrasts sharply with unselected areas of the display and the surround.

Light emission takes place at the junction of the transport conducting film and the phosphor layer. It occurs uniformly over the whole positive electrode area, and individual phosphor particles cannot be resolved with the naked eye.

The display brightness is continuously variable and the voltage range is small because of the superlinear brightness voltage characteristic. There is no threshold or striking voltage. A vast range of display formats is possible. High resolution is provided by photoetching of the substrate conductor pattern. Combinations of complex shapes both small and large may be provided on the same substrate, as may analogue and digital information together with fixed legends.


Fig. 1. Comparison of cost per bit against number of bits


Fig. 2. Assembly for a d.c. electroluminescent display
(a) Front glass. (b) Insulating mask. (c) Phosphor, which is in contact with \(g\) through apertures in \(b\). (d) Aluminium rear electrode. (e) Conductive strip. (f) Hermetically sealed cover. (g) Transparent conductive pattern (positive).


Fig. 3. Integrated circuit cost varıation with size
Fig. 4. Integrated circuit cost variation with number of pins



Complete panel styles (a) with digital speedometer; (b) with analogue speedometer


Illustration of a display designed for the Austin Princess, showing (above) all controls and (below) the presentation in the car (see also front cover).



Fig. 5. Size comparison of conventional and solid-state displays

\section*{CONSTRUCTION}

The simplicity of construction and processing are well suited to large-scale production, and flexibility in design allows economic manufacture of small quantities for prototype assessment. Custom displays are more readily realised than with any other solid-state light-emitting-display technique. D.C. electroluminescent displays are truly solid state. There is no sudden catastrophic failure mode. The slow increase in electrical resistance of the phosphor layer, which is characteristic, may be compensated for by simple constant-power-drive conditions. Neither luminous nor electrical power conversion efficiency show any appreciable decrease during operation.
The target life for this type of vehicle instrumentation is 5,000 hours. Life is measured as the time betweeen the initial brightness level down to one half of this value. Results obtained on small panels are giving encouraging results and continuous life-test data are being collected.
It is expected that such displays will first appear. on vehicles with panels of approximately \(100 \mathrm{~cm}^{2}\), containing digital clock, tachometer, fuel and temperature gauges and warning lights. This will be followed by complete dashboards containing speedometers and odometers also.

Simpler instrumentation, such as a digital clock alone, can be more suited to a digital l.e.d., or vacuum fluorescent or gas display panel. These types of displays are now available from many sources, using a standard 7 -segment digital format, and could be combined with a conventional instrument to form a clock/tachometer display for instance.

\section*{INTEGRATED-CIRCUIT TECHNOLOGY}

An instrument control system requires a combination of fairly dense logic for quartz-controlled digital clock, and odometers, linear analogue circuitry for bar-graph displays, e.g. small gauges, tachometers, etc. and display drive circuitry dependent on the type of display required. The complexity of the logic plus the fact that part of the circuit must still operate when the ignition is off will require low power. Electrical interference will require high noise immunity. Many displays require high-voltage switching.

The integrated circuits currently being designed are using integrated injection technology for the logic and high-voltage bipolar circuits for display drive. Custom-designed integrated circuits are considered best for this market because of the optimum product costing relative to the large volume requirement for the complete market.

Partitioning of these circuits has been made both to give the most flexible "kit" from which to build various panels and the lowest cost relative to Figs. 3 and 4.

As the display content grows, it becomes increasingly evident that "multiplexing" is essential to produce the lowest cost and highest reliability.' Multiplexing is the term given to controlling the display element by its two connections, in parallel with other elements, so that the minimum number of interconnections are required. A non-multiplexed panel with 256 elements will require 257 connections. Multiplexed in a \(16 \times 16\) manner only 32 connections are required.

Each connection requires a signal from an integrated circuit and a pin connection to that circuit. Typical low-cost circuits have 14 pins with a cost increase for the other standards of 16 , 18, 24, 28 and 40. Also, the cost of an integrated circuit is dictated by the area of silicon used, with a cost curve generally proportional to area up to the point where low yield increases the cost still further. Figs. 3 and 4 indicate typical variations in cost.

The set of integrated circuits that will be used is being designed jointly with Plessey microelectronics, and samples are now becoming available for complete system evaluation by various European car manufacturers.

\section*{STYLES}

The flexibility of a solid-state display to produce an attractive instrument panel is very important. The dashboard is immediately in front of the driver and the display aesthetics can be an important factor not only in the sale of the vehicle, but also in the subsequent safety aspects, and the manner in which information is transmitted from the vehicle to the driver. Early styles are shown on the cover and opposite.

As can be seen from the size information in Fig. 5, there is considerable potential space saving advantage when using solidstate displays.

\section*{FUTURE}

Further examination of trends indicates that integrated circuits are becoming more complex and larger for a given cost level, and the display technologies are also improving. The cost per bit is therefore falling for large displays, but remaining the same for small displays. From this we can deduce that it will be most economic to offer more and more facilities at the same cost level.

Microprocessors are becoming available that can compute the relationships between several input parameters and give the driver better information. The instrument panel can become the control centre for the vehicle, optimising the vehicle performance for fuel economy, braking, entertainment, etc.

Coupled with information systems, e.g. route guidance, diagnostic sensing of engine and vehicle performance, this control centre will make the driving task much easier, only giving the driver information when he requires it or in an emergency. A keyboard will probably appear, integrating calculator, and a message display panel as an aide memoire for things like shopping lists.

The boredom of a traffic jam may eventually be overcome by electronic games and in-car entertainment could take on a new meaning.

For assistance in all the above display work, acknowledgement is gratefully given to the Royal Signals Research Establishment, Phosphor Products Co., and Plessey Lid.
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\section*{munum}

Part 1-A fully protected two channel sound to light modulator custom designed for mobile work.

THIS article describes a sound to light modulator which was developed to overcome the problems found in many units. The design is intended primarily for discotheque applications, therefore emphasis is placed on reliability.

In essence a sound to light modulator converts analogue signals into a quasi-digital format which conveys the overall pattern of the music into modulation of the display lamps. An SLM can give considerable enhancement to the mainly physical class of music danced to in discotheques.

The development of SLMs, it would seem, has been overshadowed by gimmicky effects as found in random lighting displays which add little to the music, as a result they have remained relatively primitive since their inception some ten years ago.

Many discotheque operators will have experienced problems resultant of this poor state-of-the-art. These are summarised below:
(1) Continual adjustment is needed at every change in programme level in order to maintain the required degree of lamp saturation.
(2) Poor channel separation, resulting in a confused display.
(3) Unreliability, for no apparent reason.
(4) R.f.i. appearing in audio circuitry.

The circuit to be described has been designed according to the premise that simplicity is the first step towards reliability; this being an essential requirement for discotheque equipment. For this reason, the design is not totally devoid of the aforementioned shortcomings, notably the first, but is certainly a step in the right direction.

\section*{THE FAULTS DISCUSSED}

Before considering the circuit, it is useful to discuss the faults and their solutions. First (1).

A triac has a certain set voltage at which it will trigger. If the device is turned on continuously, the associated lamps will light continuously, resulting in a boring display; this situation is termed "oversaturation". When the signal level is suitably set, the triac will only switch on when most prominent parts of the music (usually the bass line and vocals) are prevalent. In this situation, the overall rhythm of the music is imitated. The associated lamps are then said to be correctly saturated. In common terms, "They flash to the music."

If the signal level is too low, the lamps will be switched on very briefly or not at all. The signal level for perfect saturation is critical, but some \(\pm 3 \mathrm{~dB}\) variation in "mean" voltage will not seriously degrade the display.

There are two ways around the problem. Firstly, the SLM could be provided with a compressor, which will ensure a constant triggering voltage. However, a simple compressor cannot discriminate between long term changes in audio level and short term changes in the dominant sounds. It will tend to bring parts of the mix which were intended to remain in the background to the forefront, and reduce the dominant rhythms to a common level, thereby giving a false display.

A careful choice of time constant can alleviate this to an extent, but complex circuitry is almost certainly involved if a complete solution is sought. Futher, setting up the SLM so that the lighting level remains constant over the anticipated range of sound pressure level requires a precise knowledge of signal levels, again requiring complex circuitry if rapid setting up is necessary. The use of a compressor was not explored because a far simpler solution exists.

If the signal for the SLM is derived prior to any variable level or tone controls, then changes in signal level resulting from the operation of these is no longer a problem. There remains the changes in level resulting from the dynamic

\section*{SPECIFICATION}

Bass - 3dB Turnover Frequency \(320 \mathrm{~Hz}-18 \mathrm{~dB} /\)
Treble-3dB Turnover Frequency Input Sensitivity Input Impedance Input Protection

Channel Power Capacity
RFI Suppression

Overcurrent Protection

Overvoltage Protection

Octave
\(3.5 \mathrm{kHz}-18 \mathrm{~dB} /\) Octave
2 V r.m.s.
\(>100\) kilohms
\(A C\) to \(>60\) volts lequivalent to 450 watts into 8 ohms)
DC to \(>90\) volts
1.4 kW (220-250 volts)
Zero voltage switched, chokecapacitor filtering
Glass fuses for small overloads to 80 amps . High speed fuses for short circuits to 850 amps
VDRs and snubbers against fuse arc and mains transient suppressors
range of the source signal (typically 50 dB for albums) and from differing modulation levels (a 12 in single having the highest, then 7 in singles, and 12 in albums with the lowest). However, since we require in any case to capture only the overall pattern of the music, herein defined as the recurrent peaks, the former point can be neglected. Regarding the latter, the 6dB leeway, previously noted, that will permit a satisfactory display is the solution to the problem of differing disc modulation levels.

Careful setting up is required nevertheless; this is speedily effected by testing the display using a noisy 12 in single and a relatively quiet album track.

It must be emphasised, however, that the direct connection of sensitive audio equipment to circuitry employing triacs provides an excellent opportunity for r.f.i. to creep into the audio chain, hence r.f.i. must be reduced to negligible levels.

\section*{POOR CHANNELSEPARATION}

Many designs feature 6 or \(12 \mathrm{~dB} /\) octave filters. Further, three channels are often used. Disco music can be adequately presented with two channels, simply bass and treble, and indeed, more channels serve only to confuse.

For two channels to act discretely, yet cover a reasonable range of frequencies,/ a high order of filter is required; \(18 \mathrm{~dB} /\) octave ( 3 rd order) filters are the compromise between filters of a higher order requiring high tolerance components and lower order filters which would make even a two channel display ambiguous.

\section*{UNRELIABILITY}

The most likely reason for unreliability is ignorance of the need for transient overload protection when triacs are used in mains circuits. For all its inherent reliability, the triac is easily destroyed, particularly when operated from the mains supply, where extremely high transient voltages and fault currents can be encountered. The incandescent lamp has two positively unwelcome characteristics:
(a) When power is initially applied to a lamp, a surge current an order of magnitude higher than the nominal current flows, reaching a peak after some \(400 \mu \mathrm{~S}\), which lasts for up to 20 mS and then dies away exponentially, reaching the nominal current level in 70 mS . In an SLM, power is constantly reapplied to the lamps, hence the surge conditions are repetitive.
(b) When a lamp filament ages, it grows brittle and will eventually fracture or "blow". The sudden change of current causes arcing across the fracture, and the heat of the arc may be sufficient to ionise the nitrogen-argon. mixture around the filament. The ionised gas will conduct heavily, resulting in a short circuit which lasts until the arc is extinguished. Filament failure in infrequent, but should it occur, an unprotected triac will almost certainly be destroyed. Fuses are the most practical means of protecting triacs from short circuit conditions in mains circuitry. Prior to considering fusing in detail, the most likely overload conditions will be identified; these are:

Small overload. Human error will make it inevitable in the long run that an SLM unit is overloaded. Given that the overload is highly unlikely to exceed 100 per cent, ordinary glass fuses will protect a triac.

Lamp surge. Fourteen 100 W lamps at 225 V will cause a 77A surge current to flow repetitively in an SLM. Any fuses used, and the triac, must be capable of withstanding this surge on a continuous basis.

Short circuit. The failure of a lamp filament can result in a short circuit and the prospective fault current is limited only by the supply impedance. Assuming worst case conditions, a current of \(1,000 \mathrm{~A}\) peak can be expected on a 30A ring circuit. Since short circuits can also occur if cables are damaged, the relative unlikelihood of filament failure causing a short circuit should not be cited as a reason for disregarding the necessity for short circuit protection.

Because semiconductor junctions can be destroyed extremely rapidly under short circuit conditions, the common glass fuses are inadequate; high speed fuses must be utilised. These have precisely defined characteristics and are capable of clearing high current faults in a matter of milliseconds.

The rapid current cut-off causes arcing and the fuse will not clear the fault until the arc is extinguished, which happens rapidly; the important point to note is that the arc voltage may be considerably greater than the mains voltage.

The fault protection arrangements in many SLMs is nonexistent or pitifully inadequate. This design is protected against the aforementioned overloads.

\section*{FUSE SELECTION}

For adequate discotheque lighting, maximum use of a 13A outlet should be aimed at. Assuming that the minimum mains voltage is 225 V , fourteen 100 W lamps per channel are permissible with enough capacity to power an amplifier and discotheque console. The lamp surge current using 100 W lamps will be 12.4 times the nominal, hence 77A at 225 V per channel.

The fuses and triacs must withstand this surge for at least 20 mS on a repetitive basis. A special anti-surge type of high speed fuse will be required. A Ferraz 600 CP-AS 10.38/12 fuse will withstand a 78 A surge for 110 mS . The same fuse has an \(I^{2}\) t of \(160 A^{2} S\) at 250 V and will clear 850A r.m.s. in 2.0 mS . A BTX-94 triac has a \(190 \mathrm{~A}^{2} \mathrm{~S} 1^{2} \mathrm{t}\) rating for this duration, and can withstand a repetitive 100A surge; hence it is suitable for the application.

Whilst the use of a 25A triac to control a 6A load smacks of gross overdesigning, it can be seen that the choice of triac is determined primarily by the available fuses, the prospective fault current and the lamp surge current.

A slow blow fuse must now be selected to protect from small overloads and to prevent catastrophic destruction of the high speed fuse which can occur when it is subjected to a steady current in excess of its nominal rating (12A in this case). Glass fuses are produced to wide tolerances, thus selection is based upon worst case samples of the fuse, and good samples may be blown. Specifically, the 5A \(1 \frac{1}{4}\) in fuse detailed in the design will handle the 6.2A nominal load current provided it is not a fast blowing specimen. In fact, the specified fuse will comfortably protect the triac up to 80A.


\section*{TRANSIENT VOLTAGE PROTECTION}

The triac must also be protected from transient voltages in excess of its rating. A compromise between an ideal yet complex transient suppression circuit lallowing low cost 400 V triacs to be used) and a very high voltage triac (needing little suppression but a lot of money) is required.

To protect the triac from the majority of transients, zinc oxide voltage dependent resistors are connected across the devices. These present a high impedance to the usual mains voltage, but present a low impedance to a transient significantly in excess of the normal voltage, thereby absorbing the energy. Fortunately, transient voltages tend to have an inverse amplitude/duration characteristic, which avoids excessive VDR dissipation. Unfortunately, the highest voltage which the VDR will reduce to a safe level is depen-
dent upon the source impedance of the transient, but in this design, transients up to several kilovolts will be clipped to below 600 V by the VDR.

Two voltage transients of known amplitude can be identified. The clearing of the high speed fuses creates a 710 V arc and the voltage rating of the chosen triac is selected to be in excess of this. The series inductor, used to mop up residual r.f.i., generates back e.m.f.s of the order of 400 V . A capacitor in parallel with the inductor acts to define this voltage.

Finally, a CR network (snubber) is placed across the triacs to assist the operation of the VDR by limiting the rate of change of voltage at the leading edge of a transient (particularly that generated by the high speed fuse) giving the VDR time to operate.

\section*{COMPONENTS . . .}
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|l|}{Resistors} \\
\hline R1 & 22k \\
\hline R2 & 1 M \\
\hline R3 & 10k \\
\hline R4 & 1k8 \\
\hline R5-R7 & 10k \\
\hline R8 & 22k \\
\hline R9 & 470k \\
\hline R10 & 22k \\
\hline R11 & 1 k \\
\hline R12 & 10k \\
\hline R13 & 3k9 \\
\hline R14 & 150k \\
\hline R15 & 22k \\
\hline R16 & 22k \\
\hline R17 & 470k \\
\hline R18 & 1 k \\
\hline R19 & 10k \\
\hline R20 & 1 k \\
\hline R21 & 10 k \\
\hline R22 & 10k \\
\hline R23 & 4710 W \\
\hline R24 & 4710 W \\
\hline R25 & Voltage dependent resistor (R.S. 238-457) \\
\hline R26 & 100 \\
\hline R27 & Voltage dependent resistor (R.S. 238-457) \\
\hline R28 & 68 \\
\hline 29 & Voltage dependent resistor (R.S. 238-457) \\
\hline R30 & 68 \\
\hline \multicolumn{2}{|l|}{All \(\frac{1}{6}\) W carbon film 5\% except where otherwise stated} \\
\hline
\end{tabular}

\section*{Capacitors}
\begin{tabular}{|c|c|}
\hline C1 & 470n polycarbonate \\
\hline C2 & 3 n 3 \\
\hline C3 & 68 n \\
\hline C4 & 180 n \\
\hline C5 & 10 n \\
\hline C6 & 10 \(\mu\) \\
\hline C7 & 3n3 \\
\hline C8 & 3n3 \\
\hline C9 & 100 n \\
\hline C10 & 220n \\
\hline C11 & 100n 1,000V \\
\hline C12 & \(100 \mathrm{n} 1,000 \mathrm{~V}\) \\
\hline C13 & \(220 \mathrm{n} \mathrm{1,000V}\) \\
\hline C14 & 100 n 250 V \\
\hline C15-C16 & 2,200 \(\mu 40 \mathrm{~V}\) elect \\
\hline C17-C18 & 100n \\
\hline C19-C20 & 470n \\
\hline C21 & \(220 \mathrm{n} 1,000 \mathrm{~V}\) \\
\hline C22 & 100 n 1.000 V \\
\hline C23-C24 & 100 n \\
\hline C25-C26 & \(250 \mu 25 \mathrm{~V}\) elect \\
\hline
\end{tabular}

\section*{Potentiometers}

VR1-VR2 \(10 \mathrm{k} \log\)

\section*{Semiconductors}
\begin{tabular}{ll} 
IC1-IC5 & 741C \\
TR1 & BC184 \\
TR2-TR3 & BC337 \\
CSR1-CSR2 & C106D \\
CSR3-CSR4 & BTX94-600U \\
D1-D2 & BZY88C-12V \\
D3-D4 & 1N914 \\
D5 & OA91 \\
D6 & BZY88C-6V2 \\
D7 & OA91 \\
D8 & BZY88C-6V2 \\
D9-D13 & 1N5401 \\
D14 & BY206 \\
IC6 & \(+15 \mathrm{~V}, 1\) A regulator, 7815 \\
IC7 & -15V, 1A regulator, 7915
\end{tabular}

\section*{Inductors}

L1, L2

T1, T2
T3

500 turns of 19 s.w.g. enamelled copper wire layer wound on \(200 \times 10 \mathrm{~mm}\) ferrite rod
1:1pulse transformer (R.S. 196-369) (Maplin)
Mains primary: 17-0-17V 1 A secondary

\section*{Miscellaneous}

LP1-LP4 Mains neons
FS1, 3, 5, 7 Slow fuses 5 A \(1 \frac{1}{4}\) in
FS2, 4, 6, 8 Fast fuses Ferraz type 600-CP-AS 10.3812A
500 mA 1 t in
\(1 \frac{1}{4}\) in panel fuseholders
Ferraz fuseholders type SGR10
S1, S4, S6 S.p.s.t. 15 A toggle
S2 D.p.s.t. 15A toggle
S3, S5
D.p.d.t. \(15 A\) toggle
\(19^{\circ} \mathrm{C} /\) watt heatsinks (2 off)
\(2 \cdot 1^{\circ} \mathrm{C}\) /watt heatsinks (3 off)


Fig. 1. Block diagram of unit and signal derivation points from typical disco console

\section*{R.F.I. APPEARING IN THE AUDIO CIRCUITRY}

The use of zero voltage switching is a foregone conclusion when sensitive audio circuitry is directly connected to and in close proximity to a potential source of r.f.i. This is totally eliminated when the triacs are triggered at the zero point of the mains cycle. This is an ideal that is difficult to achieve for two reasons-first the mains voltage is at zero for a very small period of time. Electronic circuitry has a finite operating time. And second, triacs have a minimum holding current below which they will not remain in their on state after removal of the gate trigger signal. For a typical \(I_{H}\) value of 60 mA , this implies that the mains voltage for a \(1,000 \mathrm{~W}\) load on a 250 V supply must be above 3.7 V for reliable triggering to occur. Even at this voltage, not insignificant r.f.i. will be generated. Additional filtering is therefore required.

A series inductor of several hundred microhenries will limit \(\mathrm{dl} / \mathrm{dt}\) to around \(1 \mathrm{~V} / \mu \mathrm{S}\) (with the nominal load) which is considered satisfactory for reducing r.f.i. on nearby AM broadcast receivers to negligible levels. The addition of a parallel capacitor creates a \(-12 \mathrm{~dB} /\) octave low pass filter with a -3 dB breakpoint of around 35 kHz .

If the unit is housed in a steel, or better still aluminium case, connected to a low impedance earth, and normal precautions are taken with respect to avoiding interaction between conductors carrying small signals and load currents, then to all intents and purposes, r.f.i., whilst undoubtedly occurring, can be disregarded.

\section*{INPUT AMPLIFIERS}

In order to minimise r.f.i. pickup, the input sensitivity of the SLM is set at around 2 V . For reasonably constant lamp saturation, the input signal must be derived prior to volume, tone or other equalisation controls on the discotheque console, mixer, etc. (Fig. 1). Here an input amplifier will normally
be required, Figs. \(2 \mathrm{a}, 2 \mathrm{~b}\) and 2 c . If tone or level controls precede equalisation stages, then the input amplifier must be provided with its own equalisation.

In discotheque applications, the microphone may also be required to feed the SLM, and in the case of a stereo desk, left and right channels must be mixed, Fig. 2c.

Power for the op-amps may be derived from power rails in the desk, a separate p.s.u. based on that illustrated in Fig. 4 may be provided, or the input amplifier may be powered from the \(S L M \pm 15 \mathrm{~V}\) rails.

As a last resort, the input amplifier may be mounted inside the SLM, but particular attention must be paid to screening, which should be copper based for maximum effect at high frequencies. The source impedance of stages driving the input amplifier should be less than 20 kilohms.

If the aforementioned signal derivation is too troublesome, a 10 kilohm log pot can be wired across Fig. 3 input and OV , the wiper connected to C 1 . The unit will then operate directly from speaker lines at powers of 10 to 450 W into 8 ohms and half and double these powers for 16 and 4 ohm systems respectively. The SLM input is ground referenced, so care must be taken to ensure that speaker lines are correctly polarised.

The input is protected up to 60 V ; below this signal level excessive attenuator settings are not harmful. With this method of signal derivation the lamp saturation is now subject to changes in volume and tone control settings, but its simplicity is expedient and it makes the SLM more versatile.

\section*{SIGNAL PROCESSING CIRCUITRY}

Referring to Fig. 3, C1 provides d.c. isolation. Zeners D1, D2 and diodes D3, D4 protect the unit from excessive signal input voltages, up to some 60 V , in conjunction with C 1 and R1.

(b)
\begin{tabular}{|l|l|l|}
\hline \multicolumn{3}{|c|}{ TABLE } \\
\multicolumn{2}{|c|}{ MONO SOURCE } \\
\hline\(V_{\mathrm{in}}\) & \(\mathrm{Rw}_{\mathrm{w}}\) & CCT \\
\hline 20 mV & 1 k & (b) \\
50 mV & 2 k 7 & (b) \\
100 mV & 4 k 7 & © \\
250 mV & 12 k & (C) \\
500 mV & 22 k & © \\
\(1 V\) & 47 k & (C) \\
\hline
\end{tabular}

Figs. 2(a)-(b). Input amplifiers for mono music source


Fig. 2(c). Input amplifiers for multisource inputs

Switch S1 is provided to control earthing arrangements. In certain circumstances r.f.i. may be injected into sensitive audio stages, and in most cases the switch will be open to prevent earth loops. If the unit is ever operated from a floating source, however, earthing can be conveniently restored.

IC1 is a buffer-amplifier, providing a low source impedance for the filters and some voltage gain. The filters, built around IC2 and IC4 are bass and treble respectively. These are third order Butterworth types with unity gain in their passband. Their -3 dB turnover frequencies are 320 Hz and 3.5 kHz .

VR1 and VR2 control lamp saturation.
If saturation is seriously upset when changing from 12 in singles to albums, or synthesised music with excessive treble content is played, then a switched attenuator at this point can be a useful asset, since it allows rapid correction. The operator must then develop the habit of switching the attenuator when turntable speed is changed.

IC3 and IC5 are voltage amplifiers providing low impedance drive to the zero voltage switch. Zeners D6 and D8 provide a pulsating "strobelike" display, but experimentation with different Zeners or even their omission is suggested, as a lot depends upon the type of display, thermal lag of the lamps and the application.

If one channel is found to be insensitive, then resistor R8 or R15 may be increased. A doubling of value to 47 kilohms is recommended; the corresponding resistor R10 or R16 must then be changed to 10 kilohms in order to minimise offset voltages.

\section*{ZERO VOLTAGE SWITCH}

A single polarity mimic of the mains voltage is derived from the unsmoothed but rectified output of the power supply. This voltage is applied to the base of TR1 (Fig. 5). When this is above 0.7 V , TR1 is saturated and TR2, TR3 remain off. When the base voltage drops below 0.7 V (corresponding to a mains voltage of around 11 volts), TR 1 turns off and base drive is then applied to TR2 and TR3. The pulse width is around \(200 \mu \mathrm{~S}\) symmetrical about the zero - point.

The zero crossing pulse is applied at 100 Hz to the C106D anode. Two conditions must be met simultaneously for current to flow through the pulse transformer: (a) a triggering signal must be present at the gate of the thyristor, and (b), the mains voltage must be about the zero point.

The C106D thyristor has high gate sensitivity, allowing it to be triggered directly from an op-amp. Whilst pulse transformers may be easily hand wound, the use of a commercially manufactured component which can be relied upon implicitly for reliable isolation is an important safety factor.

\section*{MAINS CONTROL CIRCUITRY}

Since fuses usually blow at an inconvenient time, a changeover switch is provided in Fig. 4 to allow rapid reinstatement. The blown fuses are isolated by throwing the switch, allowing safe renewal. Neons LP3, LP4 will light when a fuse blows, provided a load is connected. The high speed fuses have a ceramic body, so inspection will not reveal whether they or the glass fuses only have blown. If the glass fuses are intact, then the high speed fuses are "dead" but the reverse is not necessarily so.

The snubber networks C13, R28 and C21, R30 protects the triacs from voltage transients generated by the arcing of the high speed fuses when they clear, together with the voltage dependent resistors in parallel. Protection against mains borne transients is also ensured. C12, L1 and C22, L2


Fig. 3. Signal processing circuitry


Fig. 4. Control circuitry and power supply


Fig. 5. Zero voltage switch

are placed so that stray capacitances do not hinder the effectiveness of the filtering.

The chokes must be wired as close as practical to the triacs, and the intermediate wire should be covered with ferrite beads (shown as dashed lines).

Mains cables, particularly those carrying the lamp current. should be kept well away from the signal circuitry. The trigger wires to the pulse transformer must be kept short, and should be tightly twisted or in a common sheath. The use of 25A triacs to control a 6A lamp load is undeniably expensive, and if reliability is to be sacrificed for the sake of cost, then 15A plastic triacs (BT 139-600) may be directly substituted. The high speed fuses may then be disregarded.

\section*{POWER SUPPLY}

R25, R26 and C11 protect the regulators and signal processing circuitry from mains borne and control circuitry
generated transients. Neon LP2 monitors the fuse FS9. The use of a regulated supply is essential as poor mains or transformer regulation may cause interactive effects if the op-amp's supply rails are upset. If both channels switch on simultaneously, a significant drop in mains voltage may be experienced. Also, the protection diodes D3, D4 (Fig. 3) inject supply ripple into the signal line.

Supply ripple is minimised by the use of cheap plastic regulators in place of large capacitors. The BY206 diode further suppresses this effect and allows the regulators to start under a common load.

Local decoupling is provided by 100 n capacitors, preferably ceramic types, on account of their low inductance. \(250 \mu \mathrm{~F}\) capacitors are added to cure possible low frequency instability. Their use is particularly recommended if the supply leads are more than a few inches in length.

Next Month: Construction and setting up.


\section*{by K. Lenton-Smith}

Transcriptions of a well-known organist's numbers are always interesting. In Harry Stoneham's case they present a definite challenge to brain and fingers as the score is spattered with fast semiquaver and triplet runs. Browsing in my local music store recently, I came across "Play Jazz Organ" containing twelve numbers arranged exactly as Harry Stoneham plays them. These include "Michael's Theme"-his composition used to play in the Parkinson show.

Certain of the numbers in this album have been recorded on I.p. discs in the past and for those that have an organ record collection these are "Hammond My Way" (Contour 2870112) and "Two Fellas to Follow" (Tepee TPRS 100). The transcriptions tally even to the key, which at least simplifies the task of learning to play the score. The album is distributed exclusively by Music Sales Ltd., 78 Newman Street, London W1P 3LA and costs \(£ 2.50\).

There is a tendency to think that all organs sound the same when fast Leslie is applied but there is a decided difference between the two records mentioned, the first being on Hammond and the second on Lowrey with Automatic Orchestra Control. Harry's fingers are nimble enough but A.O.C. gives an added interest to his brilliant runs.

\section*{B-3000}

New models constantly appear but the Hammond 8-3000 first shown at the NAMM Exhibition in Chicago late last year deserves a special mention. Readers with greying temples will look at our illustration

and think they recognise the Hammond B-3, which was in its heyday some two or three decades ago but is now out of production though still in demand. The fabulous Ethel Smith, some may remember, featured the B-3 in playing "Tico Tico" in a film years ago: she was present at the NAMM show and no doubt the sight of Hammond's latest model brought back memories for her.
In fact, the B-3000 is the electronic equivalent of the B-3 tonewheel organ. Even the consoles are identical at first glance, but there are important differences other than the method of tone generation. Drawbars, second and third harmonic percussion and reverse-colour preset keys are as before but Grand Piano, Electric Piano and Drawbar Sustain have been added. The new organ also has a Transposer which can shift the compass up two semitones or down four.
The purist may say that this device is not necessary, but there is no doubt that playing in \(D_{b}\) sounds brighter than the key of \(C\) : the Transposer saves the mental problems of sorting out accidentals in a different key. "Bright Wave" drawbars are available on the lower manual " \(B\) " preset, allowing the player to select five flute pitches and/or four bright voices.

\section*{IT CLICKED}

Certain new rocker tabs have been added, the most interesting being one marked "Key Clicks". This ought to bring a wry smile to the faces of constructors who have burnt gallons of midnight oil trying to eliminate key clicks-but such is the state of the art !

The Hammond's original popularity was partly due to the fact that rhythmic performers found key clicks to their liking (though erstwhile tone cabinets were designed to suppress clicks as far as possible). As the particular tab was fitted during the later stages of design, this special Hammond feature appears to be as popular as ever.

Donald Leslie heard the organ when it was on the stocks and decided that a special tone cabinet was required. The Leslie 822 was thus designed to mate with the B-3000 and provides 215 W output power-sufficient for a large audience and making the installation ideal for the professional performer.
One disadvantage of the original B-3 land indeed the C-3, A-100 and RT-3
organs) was that the preset keys required only a light touch. It was only too easy to take off the " \(B\) " preset when "smearing" the manual, so being left with no drawbars in operation-an embarrassing episode in public performance!
The new organ has a more definite preset action and it is now possible to combjne several keys simultaneously for multiple combinations.

Priced at around \(£ 5000\), the B-3000 does not have a rhythm unit, automatic chords or arpeggiator, but two full 61 -note manuals and a 25 -note pedal clavier make it suitable for playing classical and light music. It makes a pleasant change to be able to report that this is a real organ with the minimum of frills, ideally suited to the professional musician.

\section*{TAPE-FOR GOOD MEASURE}

Many constructors never manage to learn to play the instruments they make and conversely others that can play go aground with electronic problems. Even so, there are those who are fortunately accomplished in both fields.

Whatever the level of musical ability, the best critic is oneself-heard on a tape playback. This method of self-discipline is excellent for-as the reader has probably found-the simple act of pressing the "Record" button will be followed by a number of disasters! The player Is under pressure, as he would be if suddenly confronted with a public performance, but at least the tape can be scrapped: better still, the tape is used to spur the player to eradicate the problems.

These points came to the fore recently when talking to a keen young musician who agreed, but pointed out that purchasing his instrument had not left much spare cash for luxuries such as good stereo pecording equipment. Having ascertained that he had a stereo amplifier and might be prepared to spend some \(£ 30\), I decided to work on the project. The following notes may be of interest to anyone in a similar position.

\section*{INSERTS}

The problems with home recording are not only associated with musical ability: there is also the question of clear, undistorted recording.
Our young friend lives close to both TV transmitters in London, so r.f. breakthrough had to be considered. Consquently, a pair of Grundig Electret microphone inserts were purchased for some \(£ 3\). They were found to give good frequency response and did not invite a background of TV sound while in operation.

The Electret inserts have an integral FET preamplifier which requires a positive supply of between 3 V and 6 V , easily found from mercury cells as the current drain is small. The insert, slide switch and cells were mounted in a 6 in . length of metal tube, the completed microphones being fitted with a two yard coaxial cable and miniature jackplug. Results from these microphones were far superior to the usual "dynamic" cassette types.

\section*{CASSETTE}

Both inserts and cassette mechanisms appear in the advertisements in this magazine. Having studied the specifications, a mechanism with mated electronics, twin VU meters, high frequency erase, auto stop and \(\mathrm{CRO}_{2}\) switch was purchased.

Having too many pending jobs already, I was pleased that the work entailed (other than a cabinet) took very little time. This involved fitting a mains transformer and two other components and arranging input and
output connectors for the stereo amplifier. Suitable series resistors for the high level input were found to be 100 k .

The overlay panel supplied was somewhat larger than necessary and was sawn down slightly and fitted with a mains switch and pilot before it was attached to the mechanism and boxed. There was one problem with some brass studs moulded in to the underside of the top panel which were 3BA, so that \(4 B A\) nuts had to be tapped accordingly.

The result was a spendid little machine fully equal to expensive counterparts. After experimenting with various microphone positions, we obtained extremely clean and accurate recordings. Given good tapeSuperferric or \(\mathrm{CRO}_{2}\)-there was no noticeable hiss from the i.c. circuitry, which is non-Dolby, of course. The combination of high-quality Japanese mechanism and Electret inserts should satisfy the limited budget reader. Does anyone want a quarter-track reel to reel recorder cheap?


CAR RADIO A.G.C.


Philips in BP 1522421 patent an interesting new approach to automatic volume control for a car radio. The patent was applied for in 1975 and is granted under the old laws. As any driver knows, when vehicle speed increases, so ambient noise also increases. It thus becomes necessary to increase the radio volume to maintain audibility and programme enjoyment.

Many types of automatic speedcompensated volume controls have been proposed in which a gain control signal is derived from a sensor which monitors the engine speed. Such systems require modification and sophistication of the car electric system.

Philips aim to simplify installation by making the audio amplifier gain dependent on the output of an electronic detector which senses ignition pulse frequency.

In car radio 12 a frequency detector 18 is coupled to the car d.c. power supply and to an audio control amplifier 21 . The supply line carries d.c. and ignition interference pulses; capacitor 26 prevents the d.c. reaching chassis and inductance coil 25 presents a large impedance to the pulses so that they pass to the frequency detector 18.

The detector has a threshold which is continuously readjusted to the value of the last pulse having the largest amplitude and produces a d.c. voltage of which the magnitude represents the frequency of the pulses. This d.c. voltage is fed to the control input of amplifier 21.

The car driver manually sets the car radio volume to a preferred level by means of a conventional control pot 22. Thereafter the signal derived at 18 from the ignition pulse rate, and thus from the engine r.p.m., controls the amplification factor at 21 to vary the audio signal level above and below the mean value set at pot 22 .

In this way an increase in engine speed and ambient noise produces a compensatory increase in audio level and a drop in speed produces an audio reduction. This produces the subjective effect of constant audio level. The patent gives details of the frequency detector circuit, complete with suggested values for all components.

\section*{DELAYED COURTESY LIGHT}

Lucas Electrical in BP 1513127 Igranted on a 1974 application under the old laws) patent a simple modification of a car electrical system which keeps the courtesy light illuminated for a predetermined period after the door is closed-but only until the engine is started. Normally, of course, the lamp is switched directly on and off by opening and closing of the door.

When a door is opened, one of two switches 18,19 closes and capacitor 35
charges through resistors 34,42 , diode 17 and the closed switch. Transistor 23 is turned on by current flow through resistors 22, 21, 42 and diode 17. Transistors 26, 28 and 31 also turn on and the courtesy lamp 37 is illuminated via a manual switch 36 , resistor 32 and transistor 31 .


When the car doors are closed, so thiat switches 18 and 19 are open, capacitor 35 discharges through resistor 21 and the base-emitter of transistor 23 to hold the transistor 23 on for a limited period of time, e.g. eight seconds. During this period transistors \(26,28,31\) are held on and the courtesy lamp remains lit.

If, after the doors have been closed and the switches 18,19 opened, the car ignition switch 14 is closed (to start the car) the capacitor 35 discharges rapidly through diode 15 and resistor 16. The courtesy lamp 37 thus goes out almost immediately.

In this way there is no risk of the courtesy lamp remaining lit while the car is driven off.

\title{
Semiconductor UPDATITE FEATURNF: TDA 1028, TDA 1029 MC 3423
}

\section*{A.F.SWITCH}

Ever had the problem of switching one of several audio signals to a common output? Traditionally, you would use a few feet of screened cable and a panel mounted switch to do the job. Of course, you would have to use a high quality switch and be very careful with your cable routing, but if the extended signal paths didn't increase crosstalk and mains hum, pick-up too much, the mechanical switch solution might be quite acceptable.
More recently electronic switching has become possible. Using this method a mechanical switch is still needed, but now it carries not the signal, but what are in effect logic levels to control the electronic "business end". You could throw one of these together with a handful of op amps and an analogue multiplexer, or you may be able to lay your hands on some op umps whose outputs can be gated into a high impedance state so that the multiplexer is not needed. In either case the solution may be rather expensive and will involve several packages.
To solve the problems of both mechanical switches and the multi-package electronic switch, Mullard have introduced a couple of devices which they call "Signal Sources Switches". The TDA 1028 and the TDA 1029 as they have been coded, look ideal for use in audio, even hi-fidelity audio, systems. They contain, in single 16 pin di.i.l. packages, complete electronic switching systems which run from single 6-23V supplies and feature input protection by clamp diodes together with full output short circuit protection. The TDA 1028 consists of two, two-pole, two-way switches, which means that it has eight input lines and four output lines which are controlled in pairs by two "Switch Control" inputs. The TDA 1029 consists of one, two pole, four-way switch, so it has eight inputs and two outputs controlled by three "Switch Control" input lines. (A fourth control line is not needed because with the three control lines floating, channel one is automatically selected.)
The typical specifications for these devices look good. Over the 20 Hz to 20 kHz band, the r.m.s. output noise voltage is only \(5 \mu \mathrm{~V}\), and a 5 V input signal in the same band suffers only 0.04 per cent total distortion. Amplitude response across the band is within 0.1 dB , and at 1 kHz crosstalk? between inputs is 75 dB down.

Apart from use as stereo quad switches, these devices would seem ideal for use in disco systems and synthesisers etc.


\section*{CROWBAR}

Ever had the experience of a voltage spike on the supply rail which ruins your delicate i.c.s? What you need is a crowbar. (No, not to smash up the errant power supply you idiot!

A crowbar is usually formed by using an SCR connected between the raw d.c. supply and ground. When an overvoltage condition is sensed, the crowbar SCR is triggered and the resulting short circuit causes an immediate drop in the voltage supply, followed by a blown fuse within a few tens of milliseconds for a more permanent solution!

The SCR part of a crowbar is simple. The difficult bit is sensing an overvoltage and generating the necessary trigger pulse, two jobs done very well by a new device from Texas Instruments, coded MC 3423.

The MC 3423 lives in an 8 pin mini-chip package and contains a voltage reference and a comparator, together with output drive transistors for (a) the SCR gate and (b) an indicator lamp or l.e.d. which must be fed from a separate supply.

The indicator is necessary because it shows the unlucky user just what has gone wrong. Without it, he would just have to suck his thumb while the wisp of smoke from the fuse drifted skywards!
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\section*{SPECIFICATIONS}
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\section*{GETTING HOOKED}

This micro hook is ideal for test hook ups to i.c.s. Weighing less than one gram it will not damage an i.c. leg while remaining tenaciously attached.


Overall length of the hook is 1.75 in and it operates with a "hypodermic" action.

Available ex-stock in ten colours the price is 62 p plus VAT and p\&p. Cash with order, otherwise minimum order charge \(£ 10\). Trade enquiries invited.
Special Products Division, British Central Electrical Co. Lid., Briticent House, New Street, Ringwood, Hants. (04254 4611 ).

\section*{FOR MAINLINE READ ARROW}

Arrow Electronics, makers of the "Leader" range of kits, have acquired the entire stock of Mainline Electronics, previously suppliers of amplifiers, audio equipment spares, semiconductors and components.

It is thought that production of the Mainline amplifiers will be continued and merged with Arrow's existing range of "Leader" kits.

Arrow Electronics Ltd., Leader House, Coptfold Road, Brentwood, Essex CM14 4BN ( 0277 219435).

\section*{PRETTY POLY}

A new live performance synthesiser employing patch switching has just been introduced by EMS. Named the Polysynthi this colourful instrument has a four octave keyboard but can cover nine octaves in six overlapping ranges. Tuning is claimed as completely stable.

Outputs from the instrument are polyphonic, pressure dependent control and position dependent control. The first control relates to pressure applied to the keyboard and the second to the highest note played. Their voltages can be routed to external devices. In the photograph the left two panels control the envelope follower, two voltage controlled I.f. oscillators and two ADSR envelope

generators. The centre panel provides the sound source; pulse, square and triangle; noise generator and external input. These can be mixed in any proportion. The two right-hand sections contain the treatment with two and four pole filters and a v.c.a. Included is an analogue delay line with voltage control of delay and with variable feedback and mix. The analogue delay line can be used for echo, reverberation, chorus and flanging. The instrument is expected to retail for around \(£ 1,000\).

Electronic Music Studios (London) Ltd., The Priory, Great Milton, Oxford.

\section*{DIGITAL BREADBOARD}

The Powerace 102 is a complete circuit building system that ailows prototyping and experimentation without the use of a soldering iron. With a regulated 5 V power supply built in , the unit has been designed specifically for digital work. Two breadboard strips allow i.c.s and/or discrete components to be plugged in directly. Both strips have a matrix of 256 terminals each with five interconnected contacts.

The distribution system comprises 16 buses each consisting of 25 electrically connected contacts. These buses may be patched around

or jumpered together in any combination to provide unique or common functions, such as voltage and ground lines, reset lines, clock lines and shift command.

Other built-in features include pulse detection ( 100 ns high or low selection); two logic switches with debounce circuitry; four data switches (logic 1 or 0 ); logic indicators ( 3 buffered l.e.d.s); clock generator offering 1 Hz , \(10 \mathrm{~Hz}, 100 \mathrm{~Hz}, 1 \mathrm{kHz}, 10 \mathrm{kHz}\) and 100 kHz \(\mathrm{O} / \mathrm{Ps}_{\mathrm{s}}\); and/or one-shot ( \(7 \mathrm{~ms}, \mathrm{Q}\) or \(\overline{\mathrm{Q}}\) )-both \(Q\) and \(\bar{Q}\) can source \(400 \mu \mathrm{~A}\) and sink 15 mA .

The price of the Powerace 102 is \(£ 105\) inc. VAT and p\&p. For further details contact Lektrokit Limited, Sutton Industrial Park, London Road, Earley, Reading, Berks RG6 IAZ (0734 669116).

\section*{THERMOCOUPLE PROBE KIT}

This kit by Comark comprises-a surface probe for medium to large surfaces, with a spring loaded tip enclosed in a ceramic sheath to minimise air current errors, measuring with good repeatability to \(+750^{\circ} \mathrm{C}\); two fast response bead thermocouples with an operating range from \(-100^{\circ} \mathrm{C}\) to \(+250^{\circ} \mathrm{C}\); a needle probe handling from \(-100^{\circ} \mathrm{C}\) to \(+600^{\circ} \mathrm{C}\); a thermocouple adaptor lead; a two metre extension lead; a quantity of thermally conductive compound packed in a syringe; and a supply of self-adhesive pads.


Price inc. of VAT and carriage is \(£ 53.46\).
Comark Electronics Ltd., Brookside Avenue, Rustington, Sussex BN16 3LF (09062 71911).


Two hundred and six stands; a warm dry atmosphere, and beer at \(90 p\) a pint, that was the prospect on entering the Grosvenor House Hotel, Park Lane, on the 27th of February. The exhibition was sponsored by New Electronics and well organised by the Evan Steadman Communications Group.

\section*{SMART CASES RULE-OK}

Proven in use in the U.S. of A. these project cases by OK Machine and Tool are what any self respecting project builder should be treating himself (herself) to as soon as a circuit idea has been breadboarded and a layout size

is known. They are available in several widths, heights, and depths in beige, blue or black. The 3 mm ABS mouldings incorporate vertical circuit board guides in 11 places, and handles and tilt leg stands are options.

Pac Tec "C" series, from \(£ 5.33\) by OK Machine \& Tool (UK) Ltd., 48a The Avenue, Southampton, Hants SO1 2SY (0703 38966). Trade enquiries invited.

\section*{NO SHORTS}

With a smart case to house your project what better next than to lash out on a completely new pair of pliers with an off-cut retaining device specially designed to prevent the inevitable short in the hidden. areas


The Swedish company Bahco make a useful range of high quality cutters and pliers, and the 2131 electronics pliers shown are the result of a detailed international survey of industrial users requirements. They have a new design of return spring and the handles are nice and wide. The jaws are available single or double bevelled. These cutters retail at \(£ 6.38\) plus VAT and Bahco tools are widely available.


\section*{TOGGLES, ROCKERS 'N' LEVERS}

Pye Electro-Devices have introduced a range of miniature switches; toggle, rocker and lever switches with conventional terminals, wire-wrap or p.c.b. mountable terminals; p.c.b. mountable toggle switches with terminals bent at \(90^{\circ}\); p.c.b. mountable toggle, rocker and lever switches with support brackets. Pye are also making an offer you cannot refuse-a sample pack of five different switches at a special introductory price of \(£ 2\) post free. With a number of types giving thousands of combinations they are obviously not able to select five switches of your choice but the pack has been made up of the most commonly used types and includes one rocker, one lever, and one p.c.b. mountable with edge actuation. This is a limited offer so check a vailability by phone first. No order required.

Pye Electro-Devices Ltd., Exning Road, Newmarket, Suffolk CB8 0AX (0638 5161).


\section*{SUPPORT YOUR P.C.B.s}

Are you still fiddling about with nuts, bolts and spacers to support your p.c.b.s and running into the snag of having to avoid any track being close to mounting metal? Nylon circuit board supports are the answer-quick to fix, boards easily lifted off for fault finding, no chance of shorts.

From Chicago, the Richco Plastic Company manufactures fasteners for industry and their stock includes a range of p.c.b. supports; spacers; circuit board guides; wire saddles; cable and wiring accessories; round and flat cable clamps; clips; bushings; extrusions; nylon rivets; circuit board ejectors and pullers.

Mainly a supplier to industry, with a \(£ 10\) min . order, you may be able to get samples. Stand of distances are from 5 to 22 mm . Or persuade your component supplier to stock.

Richeo, Universal House, 10a Gilbert Road, Belvedere, Kent DA17 5DA (032 24 47227). Trade enquiries invited.

\section*{HOLD 'EM DOWN}

One of the most common poorly secured parts inside project boxes which arrive at the P.E. offices is the battery, often wedged in one corner with a piece of cardboard, taped to the side of the box, or even just left to rattle around. The difficulty is to hold the thing firm while still allowing it to be replaced easily. Double sided sticky pads are an ingenious quick fix solution but for bother free replacement times a properly designed mechanical device gives a professional finish.


The holders shown secure those most difficult of d.c. sources, the single cylindrical cell If more than one cell is needed the nylon holders are easily interlocked to form an array.
Cambion Electronic Products Ltd, Castleton, Sheffield S30 2WR (0433 20831). Trade enquiries invited.

\section*{WHEN TWINS ARE WELCOME}

The Weir range of Twinpack power sup plies contains three models: \(0-30 \mathrm{~V}\) (c) 1 A , \(0-30 \mathrm{~V}\) @ 2A and \(0-60 \mathrm{~V}\) @ 1A. A current limiting circuit enables the user to set the

current limit to any value from the rated minimum down to zero. Prices, \(£ 150-£ 180\).

Weir Instrumentation Ltd., Durban Road, Bognor Regis, Sussex PO22 9RW (02433 26611).

Having built your project go to the next page to find a suitable DMM for testing and
fault finding, if you're not already constructing our Autoranging DMM.

The All Electronics Show had a wide range of digital multimeters on display with Gould Electronics launching their Advance Alpha IV, a \(3 \frac{1}{2}\) digit unit using liquid crystal display. This is their first Alpha type multimeter to feature a l.c.d. and it is claimed that the battery life is increased from 8 hrs. (using the I.e.d. display) up to 400 hrs. The Alpha IV has a basic accuracy of 0.2 per cent. The large liquid crystal display gives automatic indication of decimal point position, polarity and over-range. The instrument normally operates from four "C" cells, but a mains, battery eliminator option is also available.


A total of 25 measurement ranges are provided, covering d.c. voltage measurements from 200 mV to 1 kV , a.c. voltages from 200 m V to 750 V r.m.s., d.c. and a.c. currents from \(200 \mu \mathrm{~A}\) to 2 A , and resistances from \(200 \Omega\) to \(2 \mathrm{M} \Omega\). Maximum resolution is \(100 \mu \mathrm{~V}\), 100 nA and \(100 \mathrm{~m} \mathrm{\Omega}\). The unit is supplied complete with test leads, a handbook and a set of batteries. Optional accessories include the battery eliminator unit, a high-voltage probe and a carrying case. The price of the Alpha IV is \(£ 105\) plus VAT, from Gould Instruments Division, Roebuck Road, Hainaulh, Essex 1G6 3UE (01-500 1000),

\section*{BACH-SIMPSON}

The Bach-Simpson 463 has a \(3 \frac{1}{2}\) digit \(0 \cdot 5\) in. 1.c.d. display with push button selection of all ranges and functions.

The instrument has 26 ranges, 5 d.c. up to \(1 \mathrm{kV}, 5\) a.c. up to \(600 \mathrm{~V}, 6\) resistance up to \(20 \mathrm{M} \Omega, 5\) a.c. and d.c. current ranges up to 2 A , with an accuracy of 0.2 per cent on d.c. voltage ranges.


The 463 is supplied with alkaline batteries which enable 200 hrs of continuous operation. An automatic indicator gives a warning of low battery condition. The cost of the 463 is \(£ 125\) plus VAT and carriage.

Bach-Simpson (UK) Limited, Trenant Industrial Estate, Wadebridge, Cornwall PL27 6HD (020881 2031).

\section*{\(5 \frac{1}{2}\) DIGIT}

The Datron 1057 is a high quality multimeter with a \(5 \frac{1}{2}\) digit display offering a scale length of 199,999 . This enables d.c. measurements to be made to an accuracy of 0.01 per cent.

The instrument has \(5 \mathrm{d.c}\). voltage ranges \((0-1,000 \mathrm{~V}), 4\) a.c. voltage ranges \((0-700 \mathrm{~V})\) and 5 resistance ranges \((0-20 \mathrm{M} \Omega)\). A 1 kV peak can be applied continuously to any a.c. or d.c. voltage range and up to 250 V r.m.s. on

any resistance range without damage.
The price of the 1057 is \(£ 650\) plus VAT and a current shunt set is available as an optional extra for \(£ 150\).

Datron Electronics Limited, Hurricane Way, Norwich Airport, Norwich NR6 6JB (0603 404824).

\section*{LAWTRONICS}

The Lawtronics range of LM Multimeters feature four l.e.d. models ( \(3,3 \frac{1}{2}\) and two 4 digit) and two l.c.d. ( 3 and \(3 \frac{1}{2}\) digit) models. All six models have 21 ranges and are housed in the same size case \((48 \times 68 \times 100 \mathrm{~mm})\) making them ideal for field work.


The l.e.d. models are supplied with rechargeable cells and a mains adaptor whereas with the l.c.d. models they are an optional extra. Other extras include a high voltage probe \((45 \mathrm{kV})\), leather carrying case and a panel mounted flange case to replace the tilt stand. The prices of the LM range vary from \(£ 81\) to \(£ 195\) plus VAT and p\&p.

Lawtronics Limited, 139 High Street, Edenbridge, Kent TN8 SAX (0732 865191).

\section*{ANALOGUE/DIGITAL?}

If you have ever unsuccessfully tried to decide between an analogue or digital system then the Normameter D could be your ideal choice of multimeter. This instrument combines the advantages of both the digital and analogue display in one unit.

The \(3 \frac{1}{2}\) digit display allows exact readings
to be taken whilst the analogue display shows the changing state of quantities being measured. The digital unit has a maximum settling time for d.c. quantities of 2 secs. A storage circuit is used for smoothness of operation with digits changing only after an integration period. The display ca: be fixed

and released at anytime by operation of a storage key without affecting the analogue display.

At the back of the instrument there is a 24 pole socket for connection to a pfinter. The cost of this schizoid multimeter ( \(£ 462\) plus VAT) might well help you make up your mind.

Cropico Ltd., Hampton Road, Croydon, Surrey CR9 2RU (01-684 4025).

\section*{FARNELL}

Although not at the Show, the Data Precision 935 DMM which is being distributed by Farnell is well worth including in this review. The 935 is a high performance low cost meter designed for field use with all the range and function switches mounted to one side to allow single handed operation leaving the other hand free to use the probe.

With a \(3 \frac{1}{2}\) digit 0.5 in . liquid crystal display, the 935 has 29 ranges for d.c. or a.c. voltage/current and a high ( 2.8 V ) and low ( 250 mV ) resistance range.


Calibration is guaranteed for one year and full protection for over voltage, overcurrent and high voltage transients is provided. All resistance ranges will tolerate 500 V r.m.s. or d.c. without damage or loss of accuracy. A standard PP3 battery will power the 935 for over 200 hrs or an a.c. adaptor is available as an optional extra.

The \(£ 99\) plus VAT price includes test leads, battery, instruction manual and a one year guarantee.

Farnell International Instruments Ltdo, Sandbeck Way, Wetherby, West Yorkshire LS22 4DH (0937 63541).

\section*{CANADA}

New frontiers are available to the Anik-B, the communications satellite put into orbit by the National Aeronautics Space Administration of America on behalf of neighbour Canada. The launch in December by a Delta 3914 was a first for the new self-correcting guidance system. The satellite weighs about a ton and is in a geostationary orbit. The altitude is 22,300 miles at a longitude of \(109^{\circ} \mathrm{W}\). The satellite can reach the whole of Canada from coast to coast and from the United States border to \(75^{\circ} \mathrm{N}\) latitude. Beyond this the curvature of the Earth limits the usefulness of the vehicle. It brings Resolute on Cornwallis Island within the area covered. Resolute is about 210 miles East of the north magnetic pole.

The satellite was built for Canada by RCA-Astro-Electronics and is reported to be the most powerful satellite for domestic use of telecommunications. Anik-B has 12 channels in the \(6-4 \mathrm{GHz}\) band and six in the more powerful band of \(14-12 \mathrm{GHz}\). Telesat, which is the corporation that operates the Canadian communication satellite system, paid 20 M dollars for the satellite itself and another 20M dollars for the launch facilities.

The name Anik has been used for the whole series of satellites, Anik-A in 1972 and another in 1973 have been extremely successful. The word Anik is Eskimo for brother and is aptly applied, for the system has brought together all the scattered places of the second largest country in the world. With the advent of Anik-B, specific targets can be reached, even urban areas can be included without
danger of interference. The terminal receivers can use smaller aerials (which means also cheaper) than those that were required for the first two satellites.
The channels are divided between the Canadian Department of Communications and the Canadian Broadcasting Company. The Department of Communications has leased four high frequency channels in order to test a number of educational and social service schemes. These will be beamed towards professional organisations in the cities as well as to remote areas. It seems that Canada is to implement the most ambitious and advanced programme ever attempted by the use of satellites. Health and Education, which includes the extension of such services to the Eskimo settlements in the far north, will include also the teaching of English as a second language.
The launch of Anik-B was witnessed by a contingent from the People's Republic of China. The contingent, including 35 trainee astronauts, came to inspect the launch facilities. They have already begun negotiations for a satellite communications system which would be put into orbit by shuttle in 1980. The Canadian Minister for Communications, Jeanne Suave, examined the launch facilities but admitted that she was also considering the European Space Agency's "Ariane" rocket launcher for future payloads as well as the shuttle facilities. This will probably depend largely on price requirements.

\section*{MORE ABOUT VENUS}

The unique Venus expedition with its multiple vehicles is now answering some questions, but posing others. Some of the questions were answered before the probes entered the Venusian atmosphere. A number of questions were to the point before the mission, and these were based on previous missions, both American and Russian. The amount of scientific covering was extensive when the two Nations entered the vicinity of the planet with almost simultaneous missions. Much data had been received from the previous Russian landers and those from the USA. Some of those questions are answered and more will be in the months to come.

One important result of this mission was that the probes offered detailed information about the atmosphere supplementing that already available from the earlier Russian missions. Venus has no magnetic field in the sense that the Earth has; what does exist is very small, if any at all. Yet the effect of the solar wind is similar to that on the Earth. The bow shock wave which is set up in the Earth's atmosphere when the solar wind presses on the magnetosphere, appeared also in the atmosphere on Venus. This occurred
at some 750 miles above the surface of the planet. It is clear that the solar wind induces a magnetic field in the ionosphere. The shock area is the barrier to the electrons and protons of the solar wind. The strong magnetic field was apparent at the height of 750 miles. The energy involved by the leakage of some of the particles beyond the shock wave would be sufficient to heat up the atmosphere. The temperatures involved could reach \(5,000^{\circ} \mathrm{K}\).

The Orbiter spacecraft verified previous suggestions that the level of the ionopause varies in height according to the speed of the solar wind. The first week of observations by the Orbiter indicated a falling speed of the solar wind from 300 miles/second to 155 miles \(/ \mathrm{second}\). The consequence of this was that the ionopause rose from 155 miles to 930 miles. This sort of change means that the level of the ionopause is very sensitive, so much so that if there is an eruption such as a flare on the Sun then the conditions in the atmosphere would very rapidly change

It so happened that during the first week of the orbiter monitoring there was a flare. This resulted in an increase of the solar wind speed to nearly 400 miles/ second. The ionopause was flattened and was only 155 miles above the surface of the planet. The temperature of the atmosphere has an interesting gradient. It was measured to be \(27^{\circ} \mathrm{C}\) at 155 miles down to \(-90^{\circ} \mathrm{C}\) at 60 miles then increasing to \(450^{\circ} \mathrm{C}\) at the surface.

In spite of the high temperature the presence of swamps and vegetation beloved of science fiction is not possible, for it would seem that most of any water that may have existed at an earlier date has now gone. This is deduced from the fact that very little hydrogen is now escaping, which would indicate that the loss of water has been completed. However, there is evidence that there was water in abundance but that it circulated in the atmosphere and was, by the action of ultraviolet radiation, disassociated and lost from the planet.

It is a peculiar world from the human point of view. It is not possible to see the Sun or even be sure where it is in the sky (sic). Visibility on the surface is about 1.8 miles but everything is red in colour. For the most part it would appear to be almost flat everywhere but at one place there was revealed a strip about 75 miles wide which showed an elevation of 10,000 feet. There is a tremendous amount of dust. This was revealed by the toppling of one of the probes which sent up a cloud of dust which took some minutes to settle. On the whole the terrain so far seems to be like Earth and not at all like the Moon, Mercury or Mars.

\section*{SKYLAB}

The problem of the demise of Skylab has occupied space engineers for some time now. The dangers were headlined far and wide in spite of the factual reports from NASA. The space experts have been very conscious of the hazards. The space station weighs some 85 tons and in burning up during the final descent there could be about 400 or 500 pieces. Most of them would provide no more than a very magnificent firework display. There is then left the major hazard of two possibly large pieces that would survive. These could weigh around two tons.

To reduce this hazard mission con-
trollers have adjusted the station's orbit so that there is more control over the path of re-entry when the atmospheric drag takes effect. To this end a signal was sent to Skylab to command a change in attitude. This attitude will ensure that the solar panels will continue to absorb power from the Sun. An official stated that there will be enough electrical power to make some control possible down to the last moment. This control could be in action for perhaps two orbits. By firing the small thrusters it may be possible to ensure that it is a "splash down" and thus avoid damage to property. The major part of the deoris will be harmless, it is the
large pieces that are cause for concern. Since they are potentially dangerous should they fall on land that is inhabited, every effort will be made to avoid this.

\section*{GROWTH OF SATELLITE COMMUNICATIONS}

At the end of 1968 revenues from satellite communication services had reached some 420 M dollars. The growth rate is such that the US market research organisation, Creative Strategies International, has forecast that the revenue by 1983 will reach one billion dollars. This is only part of a number of forecasts in the journal Satellite Communications.

News Briefs

\section*{COUNTDOWN}

Third International Symposium and Technical Exhibition of Electromagnetic Compatibility-May 1-3. Rotterdam. Details: Mr. R. E. Gerritson (070) 906800.
Compec Europe-May 10. Centre International Rogier, Brussels. Details: 01-261-8437/8.
Welsh Amateur Radio, TV, Electronics and Computer Exhibition.Sunday, May 20. Barry Memorial Hall. There will be trade, and bring-and-buy stands. Further details: Reg Rowles on 0222-565656.
Midlands Breadboard-May 23-26. Bingley Hall, Birmingham. Details: Trident.
Intel Fair-June 11, 1979. Wembley Conference Centre. Details to be announced.
Transducer 79 + Testmex 79-June 19-21. Wembley Conference Centre. Details: Trident.
Great British Electronics Bazaar-June 28-29, 1979. Alexandra Palace, London. Details: 079922612.
1979 Microcomputer Show (incorporating the DIY Computer Fair) July 5-7. Bloomsbury Centre Hotel. London. Will include seminars. Further details: Online Conferences Ltd. Tel: Uxbridge (0895) 39262.
Harrogate International Festival of Sound-August 18-19 (public), August 20-21 (trade) 1979. The Exhibition Centre + hotels. Details: Exhibition and Conference Services Ltd., Tel. 042362677.

Telecom '79-Sept. 20-26. Palais des Expositions, Geneve. Details: Secretariat Telecom '79, Orgexpo, 18 Quai Emest-Ansermet, Case Postale 65, CH-1211, Geneve 4 (Suisse).
Eltro Hobby '79-October 3-7. Killesberg Exhibition Grounds, Stuttgart. Details: 01-236-0911.

Compec-November 6-8, 1979. Grand Hall Olympia, London. Details: Iliffe Promotions Lid. Tel: 01-261-8437/8

Electronics 79-November 20-23. Olympia, London. Details: 021705 6707.

Breadboard 79-December 4-8. Royal Horticultural Halls, Westminster. Details: Trident International Exhibitions. Tel. 0822 4671.

The next All-Electronics show will take place over the period April 29-May 1, 1980. Grosvenor House will again be the venue. Organisers are The Evan Steadman Communication Group. Details: 0799-22612.

\section*{HIGHER SPEED, LOWER POWER TTL}

TWO new third generation series of computer logic i.c.s have been announced by Texas Instruments Incorporated. They are advanced Schottley series SN74AS and advanced low power Schottley series SN74ALS. The high speed advanced series are designed for computer applications and systems that achieve efficiency through high speed switching capability. The SN74ALS series are intended for minicomputers, terminals and other electronic office equipment in which lower power consumption rather than high speeds improves efficiency.

Series SN74AS gates will be offered in 20 and 24 pin d.i.p. packages the latter with MSI functions. Series SN74ALS gates will appear in 14, 16,20 and 24 pin di.i.p.

Initially, the high speed circuits will combine high performance gate and MSI arithmetic function. Twenty five device types will spearhead the low speed series. Included are gates and dual D and J-K flip-flops as well as MSI circuits.

\section*{HAMS RALLY}

THE North Midlands Mobile Rally organised by The Midland Amateur Radio Society and Stoke On Trent Amateur Radio Society will take place on Sunday April 29th 1979 at Drayton Manor Park near Tamworth, Staffs. The Park is located on the A4091, which is within easy reach of the M1, M5 and M6 motorways, and is well signposted.

Visitors will be made very welcome. The rally opens at 11.30 a.m. Radio talk in stations on 2 metres and 70 centimetres.

Further details, car stickers etc. are available free on request from: Norman Gutteridge G8BHE, 68 Max Road, Quinton, Birmingham B32 ILB.

\section*{polnis Rrishlic}

\section*{MICROPRINTER (February 1979)}

The stripboard layout shows the correct pinouts for the BC2 12 transistors. The inset is wrong. Some links are missing from the stripboard, and these are: IC3 pin 4 to pin 9. IC7 pin 10 to +5 V . IC3 pin 13 to IC5 pin 5 . Junction at emitter of TR 15 to +5 V . The first link out from pin 13 on IC4 the "bendy" one) should be removed, to prevent IC4 pin 13 from connecting to IC1 pin 13.
The value of " \(R\) " should be \(10 \Omega / \frac{1}{2} \mathrm{~W}\).

\section*{SEQUENCER (April 1979)}

For continuity between both sides of the main p.c.b. it is necessary to solder the component leads on each side and to use Soldercon pins for the i.c.s.

\title{
soum
 Michael Tooley в.а. David Whitfield b.а.
}

THE sound operated switch described in this article can be used for any application where the presence of an audio frequency signal is required to provide control of an external circuit. A typical use is that of providing for automatic operation of transmitter receivers, or the start/stop control of tape recorders. In the former application the sound operated switch is inserted between the microphone and "press-totalk" rail, so that the operator's voice actuates the changeover relays. In the latter, the sound operated switch is used to provide automatic start of the tape recorder when the required input signal, such as speech, music, or even bird song, is present!

The block diagram of this system is shown in Fig. 2.

\section*{CIRCUIT DESCRIPTION}

The complete circuit of the sound operated switch is shown in Fig. 1. The incoming audio frequency signal is applied to a conventional voltage amplifier stage, IC1d, which has variable gain between 1 and approximately 50 by means of VR1. The outbut of this first stage is nominally

several hundred millivolts peak to peak, and this is then fed to a precision unity gain active fullwave rectifier formed by D1, D2, IC1a, and IC1c. The positive output excursion of this rectifier is applied to a comparator, IC1b, the threshold of which is made variable by means of VR2. The negative output excursion of the comparator is used to enable a conventional timer circuit which uses a 555 . The time constant of the timer monostable is controlled by means of VR3, and the positive going output of the timer is made available for control of external circuitry. A light emitting diode, D4, is incorporated in order to provide a visual indication of the output stage.

It is important to note that, once the input level has risen sufficiently to trigger the circuit and enable the timer, the output of the monostable will remain "high" (for a period of time dependent on VR3, R12, and C6) regardless of any subsequent reduction of input level.

In practice this means that there is a pre-determined delay in the output reverting to its original state and hence the circuit does not "drop out" at regular intervals when the input level falls during momentary pauses in speech or music. Those familiar with voice operated equipment will know that there is nothing worse than a voice operated switch that drops out at the end of each syllablel

\section*{CONSTRUCTION}

With the exception of the sockets, I.e.d.s, and VR3, all the components are mounted on a small single sided printed circuit board. The layout of the p.c.b., viewed from the copper foil side, is shown in Fig. 3. The corresponding component layout, on the top side of the board, is shown in Fig. 4. Care should be taken to ensure that all components are located correctly on the p.c.b. and the use of sockets for the two integrated circuits is recommended. The p.c.b. is mounted using four short 6BA pillars and, by virtue of its relatively small size, it may be easily fitted into almost any vacant space inside an existing piece of equipment. Alternatively, where the circuit is to be used externally, the use of a standard die-cast aluminium or plastic box is recommended. In this case, the box should be sufficiently large to accommodate an internal battery.

\section*{TESTING AND INITIAL ADJUSTMENT}

Carefully check the p.c.b. and external wiring before connecting the supply. Connect a 9 V battery (PP3, PP6 or similar), or alternatively a stabilised d.c. supply of 8 V to 12 V , perhaps from the equipment to which the circuit is fitted. Insert a milliammeter on the 100 mA range in the positive


Fig. 1. Circuit diagram

\section*{SPECIFICATION}

Maximum input sensitivity: 2 mV pp sine wave at 1 kHz into \(10 \mathrm{k} \Omega\)
Frequency response: 60 Hz to 15 kHz at -6 dB (see Fig. 5)
Output control voltage: standby \(=O \mathrm{~V}\)
triggered \(=7.5 \mathrm{~V}\)
Time delay (on reverting from triggered to standby): continuously variable from 200 ms to 2.8 s*
Attack time (on changing from standby to triggered): typically less than \(500 \mu \mathrm{~s}\)
Supply: 9 V d.c, at 7.5 mA standby, and 11 mA triggered. Alternatively an external regulated supply of between 8 V and 12 V may be used.
- Longer time delays may be obtained by increasing the value of VR3. A \(1 \mathrm{M} \Omega\) variable will, for example, give a time delay of approximately 30 s .

Fig. 2. Block diagram of Sound Operated Switch


Fig. 4. Component layout
Fig. 3. Printed circuit board track layout (actual size)



Fig. 5. Frequency response of the sound operated switch


Fig. 6. Waveforms (input \(=10 \mathrm{mV}\) pp at 1 kHz ). When the input level exceeds 30 mV , the pulse width on pin 7 decreases to zero
lead and check that the supply current is in the range 10 mA to 20 mA . Where an l.e.d. is incorporated to indicate that the supply is turned on, check also that this is illuminated. At one extreme setting of VR2, D4 should not be illuminated and at the other it should be illuminated. Carefully adjust VR2 until D4 is just extinguished. This should occur at about mid-position of VR2.

Set VR1 to maximum position (this corresponds to maximum gain) and connect a microphone or signal generator to the input. It should now be possible to trigger the circuit with an audio frequency signal of a few millivolts peak to peak.


Fig. 7. (a) Basic arrangement of external circuitry

(b) Use of a relay to control an external circuit


\section*{(c) VOX operation of a transmitter}

(d) Automatic recording system (audio connection to tape recorder not shown)

(e) Audio derived squelch for a receiver

VR2 can, if necessary, be re-adjusted for an increase in threshold sensitivity. Adjustment of VR3 should provide a variable delay in re-setting the output once the input signal has been removed. At minimum setting, VR3 should produce a delay which is hardly noticeable. At maximum setting the delay should be approximately three seconds. The state of the output of the switch is, of course, readily indicated by means of D4.

\section*{APPLICATIONS}

The basic arrangement of the external circuitry of the sound operated switch is shown in Fig. 7(a). The microphone used should be a low or medium impedance type \(\langle 300 \Omega\) to \(50 \mathrm{k} \Omega\) ). An l.e.d. may also be added, as shown, in order to give an indication that the supply is on. For many applications the use of a miniature relay is recommended, as shown in Fig. 7 (b). A 9 V reed type should prove suitable. When long microphone cables are necessary, the use of screened cable is recommended. This helps prevent spurious triggering of the switch due to hum and stray transients, especially when the circuit is used with maximum gain.

\section*{COMPONENTS LIST}

\section*{Resistors}
\begin{tabular}{ll} 
R1, R4, R12 & 10 k (3 off) \\
R2, R3, R5-R9 & 47 k (7 off) \\
R10,R11 & 100 k (2 off) \\
R13 & 2 k 2
\end{tabular}

All fixed resistors \(\frac{+}{4}\) W 5\% carbon types

\section*{Potentiometers}

VR1 \(470 \mathrm{k} \Omega\) subminiature horizontal skeleton preset VR2 \(100 \mathrm{k} \Omega\) subminiature horizontal skeleton preset VR3 \(100 \mathrm{k} \Omega\) linear carbon potentiometer

Capacitors
\begin{tabular}{ll} 
C1, C3 & 100 n polyester (2 off) \\
C2, C5 & \(10 \mu\) miniature 6 V electrolytic (2 off) \\
C4 & \(10 \mu\) miniature 12 V electrolytic \\
C6 & \(22 \mu\) miniature 12 V electrolytic \\
C7 & 10 n ceramic
\end{tabular}

\section*{Semiconductors}
\begin{tabular}{ll} 
IC1 & LM324N quad operational amplifier \\
IC2 & 555 timer (8-pin pack) \\
D1-D3 & OA47 general purpose diode (3 off) \\
D4 & l.e.d. light emitting diode
\end{tabular}

\section*{Voltage Table}
LM324N quad operational amplifier
IC2 555 timer (8-pin pack)
D4 l.e.d. light emitting diode
If debugging is necessary, use these test point voltages


An arrangement for the VOX operation of a transmitter is shown in Fig. 7(c). The transistor is saturated whenever the switch is triggered. Thus the "press-to-talk" rail is effectively earthed whenever the operator speaks and this actuates the change-over relays in the transmitter or transceiver. An automatic recording system is shown in Fig. 7(d). In this application the presence of a signal at the loudspeaker causes the tape recorder motor to be energised. The recorder will run as long as signals are present. The \(47 \mathrm{k} \Omega\) resistor connected in series with the input provides attenuation since, in this application, the sensitivity of the switch is too high and it is thus liable to become triggered by noise.

An audio derived "squelch" arrangement is shown in Fig. \(7(e)\). When there is no signal from the receiver the loudspeaker output is terminated in a \(10 \Omega\) resistor. The receiver is thus muted and background noise is eliminated. When a signal is present the loudspeaker is brought into circuit. The threshold of switching may be set by adjustment of VR1 and VR2.

\title{
MK14 RREVIIEWN
}

\author{
A.A.BERK b.sc. ph.D.
}


The MK14 must surely represent one of the most remarkable revolutions in computer technology to date. Much has been said about this little kit, both in magazine articles by reviewers (see Microbus in Feb. issue) and word of mouth by constructors-whatever I or anyone else may say, the complete computer for around \(£ 40\) is a landmark of, at present, unassailable proportions. The following, not uncritical, article pieces together my experience of the above device along with the description of the cassette interface by Science of Cambridge, and an interface to the P.E. VDU giving, all in all, a very cheap and powerful system.

THE kit comes pleasantly and well packaged as a set of resistors, capacitors, integrated circuits (without sockets), l.e.d. display, keyboard, reset switch, crystal, p.c.b. and even contains a 5 volt, 1 amp regulator. A "Micro Computer Training Manual" for the MK14 is also included.

The manual has clearly been designed with the absolute beginner in mind including, for instance, a good section on how to solder. However, it is generally agreed that absolute beginners and even beginners in either software or hardware alone, find it difficult to decipher some of the more fundamental concepts introduced in the manual.

The kit uses the National Semiconductors SC/MP II m.p.u. as its basis and the architecture and instruction set is described in the manual in just nine pages. Everything is there, but the beginner would do better to buy a full software introduction to SC/MP programming-I would recommend, for instance, Drury and Smart's "A Guide To SC/MP Programming".

\section*{MEMORY MAP}

The hardware of the MK 14 is also described in a cursory manner-even a complete memory map with addresses is lacking and for those who are contemplating using the MK 14 the memory map shown opposite may be useful.

In addition to this omission, until recently, no pin numbers were included on the circuit diagram which made it very difficult to expand in any way-I am very pleased to see this rectified on the latest versions.

The upper four address lines of the system are left encoded (they have to be demultiplexed from the Data Bus anyway) and hence 12 BITs of address are available giving 4 K of memory capability. This is the reason for the character " X " appearing in the memory map. The most significant hexadecimal character of the address must be used but may be anything-i.e. the addresses 0123 and F123, for instance, are indistinguishable.

The map indicates the hardware available in the kit. Included in the basic price of \(£ 39.95+£ 3.20\) VAT and postage etc., are 256 bytes of random access memory (using two 2111 chips). The display is a calculator type 9 -digit, 7 -segment multiplexed l.e.d. display giving an adequate read-out for simple programming. Space is provided on the p.c.b. for a further 256 bytes of RAM (two more 2111s) and an I/O device which itself contains 128 bytes of RAM.

If the above additions are made, the final system contains 640 bytes of RAM and 16 ports each of which may be separately configured as an input or an output line. "Handshaking" is also
\begin{tabular}{ll} 
XFFF & STANDARD RAM \\
XF00 & RAM I/O \\
XE00 & DISPLAY \\
XD00 & RAM I/O \\
XC00 & OPTIONAL RAM \\
XB00 & RAM I/O \\
XA00 & DISPLAY \\
X900 & RAM I/O \\
X800 & MONITOR \\
X600 & MONITOR \\
X400 & MONITOR \\
X200 & MONITOR \\
X000 & \begin{tabular}{l} 
Memory Map Table
\end{tabular}
\end{tabular}
possible via 8 of these lines for communication with an independent system of some kind-such as another MK 14, as the manual informs us.

The memory map also shows how much of the 4 K of immediate memory is lost by repeats. The 512 bytes of monitor, for instance, appears four times. This is due to incomplete memory decoding adopted to keep the cost to a minimum. However, it turns out that a small change is sufficient to free the extra monitor space and advice on this is given in the Issue IV kit.


The issue number, by the way, is found on the underneath of the p.c.b. This version explains how memory expansion can be effected to make a further \(1 \frac{1}{2} \mathrm{~K}\) of RAM available in the address locations X200 to X7FF (and not X2000 to X7FFF as incorrectly stated in the note). This would give a total of 2,176 bytes or RAM-quite respectable for the machine's size and basic cost, and excellent for the field of hardware control.

\section*{POOR FEATURES}

There are two features which let the kit down, however. They are, firstly, the keyboard-which relies on bridging p.c.b. tracks with "conductive rubber"-and secondly, the extremely long lead times on delivery of the kit. I received one in January which was ordered in the previous Summer! Others have experienced waits from 2 to .4 months. The first of these criticisms is easy enough to overcome and the second is sure to ease in time-you might even try turning up at their doorstep or buying from a distributor to speed matters. The delay does not seem to extend to servicing, I only waited about fourteen days for them to check the monitor ROMs which I thought were faulty.

The issue IV board comes with a note to the effect that a kit of metal-domed switches may be purchased from Science of Cambridge for \(£ 2\). The p.c.b. does, in fact, have an expansion edge connector next to the keyboard to add an external keypad, and though the manual gives no pin-out diagram for the connector, the tracks are easy to follow.

The kit is continually being updated. A few small mistakes in the earlier boards have been ironed out and the current p.c.b. is excellent-the only problem is in the extreme thinness of some of the tracks, and great care must be exercised not to break one of them.

One of the most important updates is the revised monitor program, available on fusible link PROM and pin compatible with the old one. This includes, among other things, an offset calculation for easier programming, single step facility (with some extra logic), and cassette tape interface control. The monitor occupies the same space as before by making some of the original routines more efficient.

\section*{CASSETTE INTERFACE}

The photograph shows the cassette interface being used; and though slow (about four characters per second) it is very cheap, easy to use and of straightforward construction. The use of cassette files makes a tremendous difference to any machine, saving considerable amounts of time and energy. The interface uses two integrated circuits and some discrete components on a



The positioning of the connector and extra components to link with the P.E. VDU
small single sided p.c.b. The only fault I found was the marking of the positive end of the diode-wrong on the p.c.b. but correct on the circuit diagram. To store on cassette, the length and starting address must be set up in three memory locations. The length takes up a single byte and hence a maximum of 256 bytes may be stored at a time. The starting location of the cassette routine is then set up on the MK 14's display and the recorder switched on. Pressing "GO" causes the block to be recorded-a l.e.d. on the cassette board flashes until the recording is complete. Taking information from the tape is a similar exercise. As the instructions implied, I found the volume control, on playback, to be very critical. The machine does work however, and several others I have talked to agree.

One of the most spectacular uses of the tape is in the direct loading of a block of information straight to a VDU screen. A set of notes can be kept on cassette and simply read back at leisure-especially if the tape has a tape counter for easy indexing. Another use would be for special screen formats stored on tape and used for particular programs to fill areas of the screen specially set aside for different functions. The programs involved would be very much simpler than normally necessary as the various lines and patterns used would not have to be generated in software - could be useful in the sort of minimal system for which the MK 14 is ideally suited.

\section*{P.E. VDU INTERFACE}

Of course a VDU or some sophisticated means of output, is essential for anything but the most basic uses of any machine, and below is described an interface to the P.E. VDU (published in October, November and December 1978).

The photograph shows that a 31 -way Vero connector socket has been mounted on the top edge of the MK 14 p.c.b. This is not connected to the expansion edge connector in that region but is bussed, via wire links beneath the p.c.b. to the data, address and


THE ABOVE IS FOR EARLIER MACHINES


EPT1
THIS IS FOR LATER MACHINES - SEE TEXT
Fig. 1. Interface arrangements to the P.E. VDU

MK14 and P.E. VDU connected up

control lines of the SC/MP chip. In addition, a 4011 CMOS chip is mounted at top right of the p.c.b. and connected as shown in Fig. 1.

A spare gate in IC16 is used to enable the VDU only when RDS or WDS is high-this prevents the address and status information on the data bus from interfering with the VDU RAM. The circuit shown enables the VDU as a 1 K block of memory replacing the top two copies of the monitor (each of 512 bytes).

To connect the interface to the MK14, "A11" must be disconnected from the enable inputs of IC2 and 3 (pin 13) and fed, along with "A 10", to the 4011 as shown. Pin 13 of IC2 and 3 is then connected to an output of the 4011 , again as in the diagram. The VDU "screen" then occupies the addresses X400 to X7FF. Later machines may have the monitor correctly decoded to lie only in the lowest 512 bytes-I have an issue \(V\) board which apparently has this modification. You can check by viewing a few addresses between X200 and X7FF to see if the monitor is there. If it is not then a simpler interface may be used which does not involve IC2 and 3. This is also shown in Fig. 1, the rest of the interface is as before.

It should be noted that no buffering was found necessary on any of the busses to drive the VDU as it presents, almost entirely, low power loads.

\section*{CONCLUSIONS}

The expansions described will produce not only a reasonably sophisticated personal computer, but also a control development system of some power. In my opinion it is this latter area which has been somewhat neglected by Science of Cambridge's marketing philosophy.

In conclusion, the MK 14 is a well thought out product nicely packaged, and easy to use with the revised monitor. Indeed, my firm opinion is that no electronics enthusiast or engineer should be without one in today's technology.


A reasonably sophisticated personal computer resulting from the use of the P.E. VDU with the MK14 and cassette interface

\title{
News Briefs
}

\section*{FINAL PHASE}
(n our January issue we gave a few details of the Lucas "How Far Can You Travel?" competition (see Electric Chariot Race). Well, the successful design stage entrants will now be battling it out on Sunday 2nd September 1979, at Donnington Park, Castle Donnington, Derby.

The competitor whose "home made" vehicle can transport him the furthest on two standard Lucas 12 V batteries along the \(\frac{3}{4}\) mile circuit, is the winner. Track marshalls will be on the lookout for barbed axles, and will "black flag" anyone seen slipstreaming behind another competitor.

\section*{FERRITE-CLAD AERIALS}

Adiscovery by Professor J. R. James and Dr A. Henderson of the Electronics Department at the Royal Military College of Science, has resulted in the possibility of more compact whip aerials operating with HF and VHF band portable radios.

The discovery is that a thin cladding of ferrite material can be as effective as a thicker cladding of pure dielectric material in reducing the height of high frequency whip aerials, whilst offering the advantage of increased input impedance.

The ferrite may be of a coating formed in situ, or it can be achieved by sliding ferrite beads onto the conductor. The present lack of low-loss ferrites at high frequencies imposes an upper limit of about 100 MHz generally, but the theoretical limit is considerably higher should new materials become available. Bulkiness limits the lower end to about 5 MHz .

Manufacturing licensing arrangements for interested parties are being dealt with by Jim Strutt, Computers, Systems and Electronics Group, of the National Research Development Corporation, Victoria Street, London.

\section*{RAPID RECALL MOVE}

ARGER premises at Wooburn Green now house Rapid Recall after a move from 9 Betterton Street, London.
The move was necessary because of the larger number of staff required to handle their franchise with the Digital Equipment Corporation, and cope with increasing sales of Intel and Intersil products.

The new headquarters are at: 6 Soho Mills, Wooburn Green, Bucks. Telephone, Boume End (06285) 27072/5 and 24961. Telex 849439. Rapid Recall's northern office at Nantwich remains the same.

\section*{NASCOM CLUB}

He International Nascom Microcomputer Club is now officially
launched with (at the time of writing) the publication of INMC News, issue one.

Further details about membership, which costs \(£ 5.00\) per annum, are available from Tony Rundle at Nascom Microcomputers, 121 High Street, Berkhamsted, Herts.

WHEN attempting to record or amplify low level sounds (such as wildlife) with low output, low impedance microphones, you are often faced with the prospect of trying to pick out the desired sounds from under a blanket of hiss. This hiss or noise usually comes from the pre-amplifier. A good low noise pre-amplifier, as described here, can make all the difference between a foggy, messy sound that is hard to listen to, and a clean professional sound that is a pleasure to hear.

Having had a lot of trouble with noisy recordings, the author decided to put eight low impedance pre-amps through their paces to find the one with the lowest noise output. Two of the pre-amps used i.c.'s; and it was one of these that came top. It uses the ZN459 i.c. which is truly a low noise i.c.-the noise output voltage being a third of that for its transistor counterparts.

\section*{THE ZN459}

The ZN459 is a versatile, wide band, low noise amplifier, operating up to 15 MHz , with a voltage gain of 60 dB . It basically consists of two amplifiers with gains of 100 and 10 ; giving \(1000(60 \mathrm{~dB})\) in all. The i.c. operates from a single supply of 6 V maximum; has an input impedance of 7 k at 10 kHz , and internal input biasing. The i.c. lends itself to high performance circuits with few external components.

\section*{CIRCUIT DESCRIPTION}

The circuit diagram of the Pre-Amp is shown in Fig. 1 and IC 1 is operated from about 5.5 V supplied from a 9 V battery, via dropper R1, and held steady by C1. C2 decouples the input, and determines the low frequency limit. Bypass capacitor C3 prevents extraneous signals from reaching the input circuitry through the supply lines and causing instability. The output is taken via R2, and decoupling capacitor C4. R2 is to prevent excessive loading on the output. In determining the high frequency limit, it was found that RC frequency selective feedback caused instability, so


Fig. 1. Circuit diagram of the Pre-Amp
an alternative approach was used. C5 is connected straight across the input and effectively bypasses r.f. Fig. 2 shows the modifications necessary for a 600 ohm input. Pin 5 is a point halfway along the internal feedback path. By partially


Fig. 2. Modification for a 600 ohm input
bypassing the a.c. feedback through C 6 and R 3 to OV , the input impedance is raised to 600 ohms. C5 is changed to 15 n to keep the same high frequency limit.

\section*{COMPONENTS ...}

\section*{Resistors}
\begin{tabular}{ll} 
R1, R2 & \(1 \mathrm{k}(2\) off \()\) \\
R3 & 27 k
\end{tabular}

\section*{Capacitors}

C1 \(100 \mu\) tant
C2 \(22 \mu\) tant
C3 220ntant
C4 \(10 \mu\) tant
C5 47n ( 200 ohm input) or 15 n ( 600 ohm input) metallised polyester film
C6 \(100 n\) tant

\section*{Semiconductors}

IC1 ZN459
Miscellaneous
Input and output sockets
On/off switch
Metal case \((80 \times 55 \times 25 \mathrm{~mm})\)
9 V battery
Battery connector.
Holder for 8 pin i.c. (if req)
Veroboard 0.1 in. matrix

\section*{CONSTRUCTION}

Construction is very straightforward and should present no problems. The Veroboard layout is shown in Fig. 3. Care should be taken to ensure correct polarity of capacitors, and

[605]
Fig. 3. Veroboard layour
that there are no accidental shorts between tracks. Put the i.c. in its socket last of all. The pre-amp should be housed in an all metal box for screening, with appropriate on/off switch, input and output sockets to suit individual require-


\section*{DEVELOPMENT SYSTEM}

T LOOKS as though bubbles are going to continue hitting the news for some time. Another contender in this memory race is Rock well, with a 256 K -bit block access device. But to help encourage initial development towards bubbles by OEMs, Rockwell have produced a \(\frac{1}{4}\) M-byte bubble development option for their System 65 microcomputer development system. It comprises one RCM650 controller with two RLM658 Linear Storage Modules.

The 6502 based development system can, with an auxiliary card cage, be expanded to the full 2 M -byte bubble memory subsystem capability.


Linear bubble memory module (Rockwell). The chip mapping PROM holds redundancy information about faulty loops which may exist within the bubble memory.
ments. The circuit board should be mounted on non metal supports to avoid short circuits. The pre-amp is now ready for testing.


\section*{TESTING}

Check there is about 5.5 V across pin 2 and OV . Current consumption should be about 3 mA . Plug in a microphone and monitoring source (with medium to high input im-pedance) and listen. The result should be a clean, distortion free, noise free sound.

\section*{LIMITATIONS}

With a maximum input of only 1 mV before clipping occurs (due to the i.c.'s input circuitry), the pre-amp would be unsuitable for use with a singer's hand-held microphone, for example. Also, the pre-amp could be powered from existing equipment, but the supply would have to be exceptionally ripple-free, if not battery operated, to give good, hum-free results.


\title{
NWIDRMEINTE WULIIMETER mar
}

\section*{ALEX KOWALEWSKI Dr. MARK A. SAWICKI}

This final part concludes with the remaining theory, and overall construction

\section*{DISPLAY BOARD}

THE Autoranging Multimeter always requires that Over and Underrange must be detected. In our particular situation only digits 1 and 2 need to be sensed.

Overrange Digits 2, 3, 4 are all "off".
Underrange Digit 2 shows 0 or 1 and Digit 1 "off". Consequently segments \(1 \mathrm{~b}, 1 \mathrm{c}\) (connected in parallel) and \(2 \mathrm{a}, \mathrm{b}\), f, d, must be interfaced so that the logic knows whether they are "on" or "off"

The complete Display Logic Interface employs five similar circuits as presented in Fig. 2.1 shaded portion.

\section*{INTERFACE CIRCUIT}

If the segment is off, point \(A\) is at +5 V , the transistor is off and \(B\) is -5 V . If on, point \(A\) falls to approximately +3 V due to the voltage drop across the l.e.d., the transistor turns on and point B rises to +5 V . The capacitors on the interfaces for segments \(b\) and \(g\) are specially provided to stop false overrange indications during the 6 to 7 transition at digit 2 .

The complete circuit diagram of the display logic interface is shown in Fig. 2.1. The decimal point and the lowest range l.e.d. indicator wiring diagram is quite straightforward.

Details of the p.c.b. layout can be found in a combined form at Fig. 2.3(a) and Fig. 2.3(b).

The display board and the Main Board are connected together using "edge inter-board connectors".

This is an economical and simple method of making rightangled connections between p.c.b.s by just inserting solder pins on the 0.1 inch p.c.b. pitch of the Display/Main Boards, i.e. plug (23-way) is soldered into the Display Board and the relevant socket is fixed to the Main Board. The DL 727 data is given in Table 2.1.

\section*{PSU}

Approaching the final part of this project, we now give details of the power supply circuit. This is shown in Fig. 2.2, and the p.c.b. layout is shown in Fig. 2.4. The system uses a pair of popular LM309K regulators and all the other things are pretty standard. A protection fuse of 200 mA on the

\section*{MAIN BOARD}


LOGIC BOARD


Fig. 2.1. Circuit diagram of Display Board

Table 2.1. DL727 dual numeric display. Viewed from display side: Digit \(1=\) left, Digit 2 = right.
\begin{tabular}{cll} 
Pin & & Designation \\
1. & - & Cath E (No. 1) \\
2 & - & Cath D (No. 1) \\
3 & - & Cath C (No. 1) \\
4 & - & d.p. Cath (No. 1) \\
5 & Cath E (No. 2) \\
6 & - & Cath D (No. 2) \\
7 & - & Cath G (No. 2)
\end{tabular}

Cath E (No. 1)
Cath D (No. 1)
d.p. Cath (No. 1)

Cath E (No. 2)
Cath G (No. 2)

Cath C (No. 21 d.p. Cath (No. 2) Cath B (No. 2)
Cath A (No. 2)
Cath F (No. 2)
Anode (No. 2)
Anode (No. 1)
Cath B (No. 1) Cath A (No. 1) Cath G (No. 1)
Cath F (No. 1)



\section*{COMPONENTS . . .}

\section*{Transistors}

TR1-TR5

Displays
X1-X2, X3-X4
D1

BC159 (5 off)

\section*{Resistors}
\(3 k 3\) (5 off)
\(4 n 7\) (2 off)

[609]


\section*{COMPONENTS . . .}

\section*{POWER SUPPLY}

Resistors
R1, R2
Capacitors
C1, C2
C3. 44
C5, C6
\(700 \mu 25 \mathrm{~V}\) (2 off)
220 n (2 off)
470 n (2 off)

Semiconductors
D1-D4, D5-D8
IC1, IC2
REC 70 (2 off) LM309K (2 off)

Transformer
T1 \(240 \mathrm{~V} / 2 \times 6 \mathrm{~V}, 0.5 \mathrm{~A}\) sec's. RS type 196-296 or equivalent (via Ace Mailtronix)

\section*{PSU BOARD}


Table 2.2. Analogue Switch Allocation

\begin{tabular}{|c|c|c|c|}
\hline 1/O pins & Control pin & Switch & Function \\
\hline & & IC1 & \\
\hline 3-4 & 5 & S3(b) & OHM \\
\hline 1-2 & 13 & S3(c) & OHM \\
\hline 10-11 & 12 & S4(a) & OHM \\
\hline 8-9 & 6 & S4(b) & OHM \\
\hline & & IC2 & \\
\hline 3-4 & 5 & S 1 (a) & AC \\
\hline 1-2 & 13 & S2 (a) & \(\overline{\mathrm{AC}}\) \\
\hline 8-9 & 6 & S2 (b) & \(\overline{\mathrm{AC}}\) \\
\hline 10-11 & 12 & S3(a) & Reset \\
\hline & & 1C3 & \\
\hline 10-11 & 12 & S1(b) & AC \\
\hline 1-2 & 13 & S1 (c) & AC \\
\hline 3-4 & 5 & S2(c) & \(\overline{A C}\) \\
\hline
\end{tabular}


REF-O2 Regulator


Fig. 2.5. (a) Main Board p.c.b.
(b) Component layout
(b)


\section*{AUTORANGING MULTIMETER}

mains Live side follows two sets of contacts of the mains switch mounted right at the bottom of the function switch. In order to help you with the function switch wiring, Fig. 2.8 and Fig. 2.9 show both the top and the bottom halves of this switch's wafers and associated wiring.

\section*{MAIN BOARD PCB DETAILS}

The Main Board p.c.b. layout is shown in combined form in Figs. 2.5(a) and 2.5(b). The reason for insertion of a few components on the copper side was to keep the layout around the 7107 device as similar as possible to Intersil recommendations. However, this does not create, in the author's opinion, any serious constructional problems.

Some additional information about component layout can be found from the photographs.

\section*{LOGIC BOARD}

The Logic Board p.c.b. is shown in Fig. 2.6.

\section*{RELAY BOARD}

The Relay Board p.c.b. layout is presented in Fig. 2.7(a) and 2.7 (b). The only components soldered into the copper side are the five 1 N4148 diodes in parallel to the reed relay coils. The reed relays chosen are the popular and economic, yet still high quality, RS348-986 type. The use of this particular type of relay is not critical as long as its rating and size are similar.

\section*{INSTRUMENT ENCLOSURE}

Since most of you will be electronically minded rather than mechanically minded, you may prefer to get hold of a ready-made attractively styled instrument case complete with stick on feet. We used the RS standard case No. 509901.

A display window was then cut in the front panel, and the rotary function switch, Mode/Auto manual and Range Down pushbutton switches with corresponding yellow Manual l.e.d. were placed on the left side. All the input sockets (continental 4 mm type) were set in at the bottom of the front panel. For ease of identification we used red, with the exception of the Ground socket, which was green.

All Autoranging Multimeter p.c.b.s and fuse holders (except the mains power supply fuse) are mounted into a specially provided aluminium chassis. Last but not least, the rotary function switch requires a little perspex plate ( 4 mm thick) to lock the device and prevent overturning the switch.

\section*{CALIBRATION}

After calibration of "Set Ref", described last month:
(1) If a known r.m.s. source is available or another accurate DMM, adjust AC Cal. until the reading agrees with the source or the other meter.
(2) Short the 10 n capacitor on function switch and C1, C2.

Preferably keeping to same range: put meter in DC Volts mode, measure any d.c. voltage. Switch to AC Volts mode and adjust AC Cal until reading is 3.2214 times the previous one in DC Mode.

Remove shorts.

\section*{IMPORTANT NOTE}

It is advisable to calibrate the instrument without removing it from its grounded case.


Fig. 2.8. (Top) Shows top half of S1
Fig. 2.9. (Above) Shows bottom half of S1



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The above photograph shows the tool dismantled. This enables the elastic band, which forms the plunger return spring, to be replaced should it ever perish. The two halves of the body can be parted by carefully inserting a blade in the joint. It is then a simple operation to replace the band and reassemble the tool.

The two halves are held by four fixed pegs which are a tight press fit, no adhesive is used on the joint. It is of course possible to use the tool without the return spring, though it will lose its smooth "feel"


\section*{Telecomms}

Nothing illustrates better the cruel dilemmas resulting from electronics than the telecommunications business. The old faithful electro-mechanical telephone exchanges are having to give way to modern electronic exchanges requiring less space and less day-to-day maintenance as well as being more efficient and flexible in use and cheaper into the bargain.

But what an upheaval! The Strowger exchange using step-by-step switching through uniselectors and relays is labourintensive in manufacture. Fine when you could employ low-cost female labour, displaced from the old textile mills, on assembly lines. But times have changed and not only are labour rates much higher today but legislation on equal pay means that the girls cost just as much as the lads. But in any case the run-down of manufacture of the old equipment means redundancy and we already have enough unemployment.

Plessey is one of the unhappy firms faced with problems of the gradual phasing out of Strowger production for the Post Office. Just to change over to electronic exchange production solves nothing because only a quarter of the workforce would be required. Either way people must go, "especially as the Plessey factory in Liverpool is said to be losing \(£ 7\) million per year. The Post Office is reported to be negotiating a revised price for Strowger equipment which will ease the financial problem for Plessey but not cure it in this particular plant. It is a problem which will continue well into the 1980s as electronics products continue their penetration of markets traditionally based on electromechanical devices.

The Strowger exchange has a curious history. Its inventor was not, as might be imagined, a highly trained scientist working in a well-equipped laboratory. Almon B. Strowger was a Chicago undertaker whose rival appeared to be getting all the orders for laying out, embalming and subsequent burial. The story goes that he attributed this not to his own lack of skill or even pricing. policy but to manual switchboard girls who
directed all enquiries to the rival company. He thought an automatic switchboard not relying on people would ensure getting his share of undertaking work and so he set about inventing one. This was in 1889. By 1912 the Strowger system was in full use and, with only minor improvements, has continued to the present day and beyond.

For although the days of Strowger are clearly numbered the British Post Office will still be taking deliveries of the system until at least 1981 and assuming a service life of 20 years it seems that uniselectors will still be clicking away in the 21 st Century. It just shows how a fundamentally sound system can last in the face of much more modern technology. The other example is the carbon granule microphone, virtually unchanged in a hundred years, still used in our telephones. Now the Post Office is searching for a modern electronics replacement, a unit which can be an exact replacement insert but with better and more consistent speech quality.

\section*{Time Scales}

We are so frequently overwhelmed with a seemingly endless flow of new inventions and ideas that they seem almost instantaneous. Actually, time scales are often much longer than we imagine. The semiconductor effect was observed in lead sulphide and iron sulphide in 1874. It was 75 years later that the first crude pointcontact transistor was made in a laboratory and another 13 years before the transistor found general acceptance in the electronics industry.

With all the present excitement about VLSI we tend to forget that the first MOS LSI circuits, albeit with a modest device count of only 2,500 elements per chip. were available ten years ago.

Some inventions have to wait for technology to catch up. Pulse code modulation (PCM) was proposed in the 1930 s but had to wait for the invention of the transistor before becoming a practical proposition. The laser was a fine invention in 1960 but remained for years a scientific toy while waiting for suitable applications to be found.

On the other hand development can be quite rapid once the principles and the learning curve are established. In little more than a decade communications satellites have increased their capacity from 240 duplex voice channels (Intelsat II, 1967) to 12,000 channels (Intelsat V, 1979).

\section*{Speed}

State-of-the-art in oscilloscopes was 500 MHz bandwidth last year. Now Tektronix has made a huge leap forward in Model 7104 with a bandwidth of 1 GHz , allowing direct observation of signals which on less sophisticated equipment are normally invisible. The writing speed of the trace is now an almost incredible \(20 \mathrm{~cm} / \mathrm{ns}\).

To get such performance Tektronix had to improve many aspects of oscilloscope design, not least the CRT itself. As it is virtually a microwave instrument, the p.c.b.s are based on microstrip techniques
and even the interconnections are based on microwave technology to reduce internal mismatches and reflection losses. Costing about \(£ 15,000\) with an appropriate range: of plug-in accessories, it is hardly designed for the impecunious hobbyist. But it will clearly appeal to the professional who can now record the fastest transients on ordinary film rather than the elaborate and cumbersome camera techniques formerly necessary.

Control Data Corporation's Cyber 203 computer, just announced, is also another record-breaker. Six times as fast as its predecessor the 203, using a method named vector processing, performs a handy 100 million calculations per second. The 203 is available in a range of sizes, the largest of which has a memory of two million 64-bit words. Price range is a little under £ \(\mathbf{~}\) million to almost \(£ 6\) million. First customer is the United States Air Force for three systems.

\section*{Computer World}

Clearly, the big mainframe computers like the Cyber 203 still command a market but today's pioneering area is still in microsystems for the hobbyist, the small business and education. The United States, trend-setting as ever, now has hundreds of computer stores doing over-the-counter business. In the UK, Commodore is following suit with a dealer network for the Pet computer which is expected to expand to - 100 retail outlets this year.

At the recent Microsystems ' 79 exhibition in London, some 60 firms showed systems costing as little as \(£ 100\) up to \(£ 5,000\) or so. Add-on sales are important. According to one market research report people are spending up to twice the initial processor cost on peripherals and other enhancements to their systems.

\section*{VLSI Stakes}

GEC-Fairchild, having now appointed its full board of executives, settled on a site at Norton, Cheshire, and with appropriate finance lined up, look like meeting their target date for production next year of general purpose chips for the mass market, as much as two years ahead of linmos, Britain's other new entrant. A strictly nononsense commercial venture. GECFairchild while expecting to employ 1,100 people by mid-1981, will have all assembly work done in the Far. East where labour costs are lower. It may seem a pity but necessary if the project is to be successful.

Incidentally. Fairchild has pulled out of the electronic watch business altogether and is cutting back on MPU-based video games. Market instability and low profit margins in the watch game are given as reasons. Fairchild made record sales in 1978, up to \(\$ 533\) million with profits doubling to \(\$ 24.8\) million.

Meanwhile, Inmos has been busy recruiting both in the UK and USA. The U.S. headquarters and technology centre are now to be at Colorado Springs. In the UK the Bristol site will be the technology centre but not necessarily the company HO.



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\section*{CHEAP THERMOMETER}


THIS inexpensive thermometer gives suprisingly good stability and accuracy despite its simplicity.

The heart of the circuit is TR1, which forms the temperature sensitive probe with its own inbuilt amplification. This stage works in the \(\mathrm{V}_{\mathrm{inf}}\) multiplier mode, common in power amplifiers, which amplifies its temperature dependent base-emitter voltage to about \(-25 \mathrm{mV} /{ }^{\circ} \mathrm{C}\). This voltage is then buffered by TR2.

VR2 picks off an offset voltage to bias TR3 such that the voltage between the emitters of TR2 and TR3 is proportional to the temperature. ME1 and VR1 form a voltmeter to measure this voltage.

TR 2 and TR 3 should be glued together for thermal tracking. The supply voltage is regulated by R8, D1 and D2. The use of a high voltage Zener diode is avoided because of its inherent temperature sensitivity. The metering unit is thus fairly free from temperature dependance enabling the probe alone to determine the meter deflection.

The probe should be connected to the metering unit via a suitable heat resistant cable. The leads of TR1 should be encapsulated in a silicon adhesive.

VR2 is adjusted so that the meter reads 0 with the probe in melting ice. VR1 is then adjusted so that the meter reads full
scale with the probe in boiling water. The meter scale will then be linear, reading 0 to \(100^{\circ} \mathrm{C}\).
P. R. Williams,

Stevenage, Herts.

\title{
4 WIRE-8 WAY BELL SYSTEM
}

WE have four offices, each associated with one of four laboratories some distance away. Each lab and office has a pair of phones on the same extension number, and my bell system is used mainly to call the phones on the other end of the 'pair'. The bell system is, of course, not electrically connected to the phones.
It can be seen that two of the four wires are the power lines and are fed (at any convenient point in the system-ours is fed in one of the labs) with an a.c. voltage to suit the chosen bells or buzzers. The other two calling lines are steered to the appropriate bells by diodes, thus each bell receives half wave rectified a.c. when its calling line is connected to the supply line via a diode of corresponding polarity.
Only two pairs of units are shown on the circuit diagram-the other two pairs are identical, but use calling line \(\mathbf{C}\). The office

units are shown with change-over push buttons-wired this way, the local bell does not ring when the push button is pressed. The lab units are shown with simple s.p.s.t. buttons-wired this way, each push would ring the bell at both ends of the line at once.

If line D is made an earth return, then the circuit becomes a 'three wire eight way' bell system!
C. P. Finn,

Southwood Park,
Beverley.

\section*{CMOS LOGIC PROBE}

THE probe employs two ' \(B\) ' type CMOS integrated circuits to provide a comprehensive logic probe together with a square-wave signal injector. Four lightemitting diodes show the state of the circuit under test. D1 indicates a logic level ' 1 ', whilst D2 indicates a logic level ' 0 '. These two are isolated from the external circuit via inverters.

Any logic pulses fire IC2, a dual monostable multivibrator. Positive pulses fire monostable ' \(a\) ' which feeds D3, whilst negative pulses fire monostable ' \(b\) ' which feeds D4. Thus a continuous wave-train will light all four. \(\$ 1\) switches between the logic test function and the signal inject function. The signal injector is formed from the two spare inverters in IC1, and oscillates at approximately 1 kHz .
D. P. Akerman, Coventry.


\section*{VARITONE STYLOPHONE}

\(\mathrm{H}^{\mathrm{E}}\)ERE IC1 is connected as an oscillator with a frequency determined by the presets. By touching the probe onto the printed circuit board's "keys" (shown in the etching detail), which are connected to their corresponding presets, IC 1 is made to give a frequency of between \(\mathbf{B}^{b} 7(3729 \mathrm{~Hz})\) and G9 \((12,544 \mathrm{~Hz})\). This squarewave signal is fed into IC2, a CMOS binary divider with its outputs connected to rotary switch S1, which selects the octave required and feeds the signal through TR2, via a volume control potentiometer. The amplified note now passes through a loudspeaker, giving a musical sound. Any number of keys may be included, providing there is a preset to tune each note.

A tremolo effect is added using the circuit around TR1, which provides a 6 Hz modulation of variable depth when fed into pin 5 of IC1. Some interesting effects may also be obtained using the "tone control" potentiometer. However, this does alter the pitch of the note produced, and so a switch is included to disconnect it in normal operation.

The squarewave tone output may be modified by passing it through an audio transformer T1 as shown, D1 being required to soak up any back e.m.f. from the windings. The modified output has a harsher quality than the plain squarewave sound.
The instrument is tuned using the fifteen 4.7 kilohm and seven 10 kilohm presets
connected in series. They should be adjusted to tune the notes starting at the highest and working down the scale.
T. J. Barker,

Reigate,
Surrey.

\(\mathrm{Bb}_{6} 7\)


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