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400V: 0.001, 0.0015, 0.0022, 0.0033 8p; 0.0047, 0.0068, 0.01, 0.015, 0.018 9p; 0.022, 0.033, 10p; 0.047, 0.068 14p; 0.1015p; 0.15, 0.22 22p; 0.33, 0.47 39p; 0.68 45p, 160V: 0.039, 0.15, 0.22 13p; 0.33, 0.47 22p; 1.5 33p; 2.2 38p; 4.7 47p, DUBILIER: 1000V: 0.01, 0.015 16p; 0.022 18p; 0.047 16p; 0.1 34p; 0.47 43p.

POLYESTER RADIAL LEAD (Values in μF) 250V:
0.01.0.015 6p; 0-022.0.027 7p; 0-033.0.047.0-068.0-1 8p; 0-15 12p;
0.22.0.33 14p; 0-47 16p; 0-68 20p; 1-0 24p; 1-5 27p; 2-2 31p.

FEED THROUGH
CAPACITORS
100μF 350V 8p

SWITCHES+ TOGGLE 2A 250V SPST DPDT

4 pole on/off 58 SUB-MIN TOGGLE SP changeo

DPDT 6 tags DPDT Centre off

SLIDE 250V:

24 27

26p 35p 58p

92p 115p

Break 25p 28p 23p

ELECTROLYTIC CAPACITORS: Axial lead type (Values are in µF) 63V 0.47 1.0 1.5 2.2 2.5 3.3 4.7 6.8 8 10 15 22 9p; 47 32 50 12p; 63 100 27p; 50V 1.07p; 50 100 220 25p; 470 50p; 1000 25p; 200 68p; 40V 22, 33 9p; 100 12p; 330 62p; 4700 64p; 35V 10.3 3 7p; 330 470 32p; 1000 48p; 25V 10, 22 47 6p; 80 100, 160 8p; 220 250 13p; 470 640 25p; 1000 27p; 1500 30p; 220 41p; 3300 52p; 4700 54p; 16V: 10, 40 47, 68 7p; 100, 125 8p; 330, 470 16p; 1000 1500 20p; 220 34p; 10V 4 100 6p; 640 10p; 1000 14p. TAG-END TYPE: 70V: 2000 98p; 4700 121p; 50V: 3000 75p; 40V: 4000 70p; 2500 65p; 25V: 4700 48p; 2000 37p; 40V: 2000 + 2000 95p; 325V: 200 + 100 + 50 + 100 190p.

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100V: 0 001 0 002 0 005 0 01µF
0 015 0 02 0 04, 0.05, 0 056µF
7p
0 1µF 0 15 0 2 9p 50V: 0 47;
11p

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MINIATURE TYPE TRIMMERS 2 5 6pf 3 10pf 10-40pF 5 25pf 5 45pf 60pf, 88pf

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COMPRESSION TRIMMERS 3 40pF 10 80pF 25-190pf 100 500pF 1250pF

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DIN 2 PIN Louds 3, 4 5 Pin A		PLUGS 13p	500	KETS 8p	In Lin	Push to ROCK	CHES * Miniature Make 15p ER (white) 10A 250	Push to Break
CO-AXIAL	(TV)	14p	_	4р	14p	ROCK	ngeover centre off ER: (black) on/off 10 ER: illuminated (whi	

| NULKER: Highmarled (White) | S2p | Lights when on: 3A 240 | ROTARY: (ADJUSTABLE STOP) | 1 pole / 2-12 | way 20/2-6W 3P/2-4W 4p/2-3W 8p double 10p 3-wey assorted colours Metal Screened DIL SOCKETS± (Low Profile - Texas) 8 pin 10p; 14 pin 12p; 16 pin 13p; 18 pin 20p; 20 pin 26p; 24 pin 30p; 28 pin 42p; 40 pin 58p.

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VOLUME 14 No. 11 JULY 1978

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Our August issue will be on sale Friday, 14 July 1978 (for details of contents see page 849)

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	0.96	5 - 35	2	6	70
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1	1-14	7-67	5	10	72
1	1 - 32	8 - 99	6	12	116
	1 - 32	10-39	8	16	17
	2.08	13-18	10	20	115
	2.08	17 - 05	15	30	187
	OA	26 - 82	30	60	226

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nection	to appropriat	e taps.	
Ref	Amps	£	PAP
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20	3.0	6 - 20	1-14
21	4.0	7-44	1-14
51	5.0	8-37	1 - 32
117	6.0	9-92	1.45
88	8.0	11 73	1 - 64
89	10.0	13-33	1 - 84

Ref	Amps	£	PAF
112	0.5	2.64	0-78
79	1.0	3.57	0.96
3	2.0	5 - 27	0.96
20	3.0	6 - 20	1-14
21	4.0	7-44	1-14
51	5.0	8-37	1 - 32
117	6.0	9-92	1 - 45
88	8.0	11 73	1 - 64
89	10.0	13.33	1 · 84

	00			0	
	50 VO 220/240V -20V or 25		20-25-33		24 ne
	to appropri	iate taps		P & P	R
102	0.5		3-41	0.78	1:
104	2.0		E. 08	1.14	12

	103	1.0	4-57	0.96
	104	2.0	6 - 98	1 - 14
	105	3.0	8-45	1 32
	106	4.0	10.70	1 - 50
	107	6.0	14 62	1 - 64
	118	8.0	17:05	2.08
	119	10.0	21 - 70	OA
9	113	,,,,		
	MAH	NS ISOLATI	NG (SCRE	ENED)
	PRIM	120/240 SEC 1	20/240 CT	TAPPED
	Ref			PAP
	07.	20	4-40	0.79
	149	60	6 - 20	0.96

		ING (SCRI	
PRIM 12		120/240 CT	
Ref	VA (Wall	s) £	P&P
07°	20	4-40	0.79
149	60	6 20	0.96
150	100	7:13	1:14
151	200	11 - 16	1 · 50
152	250	12 - 79	1.84
153	350	16 - 28	1.84
154	500	19:15	2 · 15
155	750	29 - 06	OA
156	1000	37 - 20	OA
157	1500	45 60	OA
158	2000	54-80	OA
159	3000	79 - 05	OA
*Please	specify 115	or 240V requ	uired

	Prim 200/22	LTAGE MAI DLATING 20V or 400/440	V
Va	Sec 100/1	20V or 200/240	PAP
60	243	5-89	1 - 32
350	247	14-11	1 · 84
1000	250	35.65	OA
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Ref	Amps	£	PAP
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79	1.0	3.57	0.96
3	2.0	5 - 27	0.96
20	3.0	6 - 20	1 14
21	4.0	7-44	1-14
51	5.0	8-37	1.32
117	6.0	9-92	1.45
88	8.0	11 73	1 - 64
89	10.0	13.33	1 · 84
	60 VO	TRANGE	

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Ref	Amps	£	PAP
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126	1.0	5-58	0.96
127	2.0	7-60	1-14
125	3.0	10.54	1 - 32
123	4.0	12-23	1 - 84
40	5.0	13.95	1 - 64
120	6.0	15 66	1.84
121	8.0	20 - 15	OA
122	10.0	24 - 03	OA
189	12.0	27 - 13	OA

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64 75 0-115-210-240 3-95 0-9	1
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4 150 0-115-200-220-240 5-35 0-9	6
67 500 0-115-200-220-240 10-99 1-6	4
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93 1500 0-115-200-220-240 23·36 OA	
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	ble in & 115\		
VA	£	PAP	Ref
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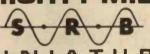
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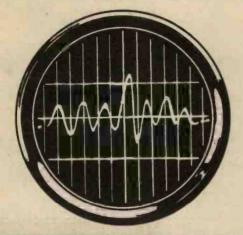
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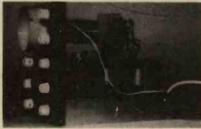
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93n		zeners	from	4V3 to	
OA8	15p;	OA91,	8p; 1	N4148.	4p.
SCR 0.8A	S	60V	T	092	35p
1A		400V	T	05 0220	60p
4A 4A		200V 400V	T	0220	52p 70p
6A		200V	T(0220	56p
6A 6A		400V 400V	T	0220 0 66	75p 80p
10A		100V	T	0220	82p
10A 10A		200V 400V	T	0220 0220 0220 0220	87p 120p
10A		600V	T	0220	148p
Triac	S	400V			98p
6A 8A		600V	T	0220	135p
15A		200V	T	0220	135p
15A		400V		ORS	220p
AC127	18p	BC548	10p	BRY56	40p
AC128 AC178	18p 18p	8C549 8CY70	10p 15p	OCP71 TIP41A	120p 56p
AC187	20p	BCY71	15p	TIP42A	66p
AC188	20p	BCY72	14p	TIP 2955	86p
AD149 AD161	70p 40p	B0131 B0132	38p 40p	TIP3055 TIS43	42p 35p
AD162	40p	BD133	48p	2N2646	6 Op
AF279	75p	BD137	40p	2N2905	21p
8C107 8C108	12p 10p	BD138 BD139	40p 42p	2N2926 2N3053	12p 28p
BC1080	12p	BD140	44p	2N3054	52p
BC109 BC1090	12p	BF173 BF181	20p 30p	2N3055 2N3442	50p
BC147	10p	BF194	10p	2N3702	130p
BC148	10p	BF195	10p	2N3703	10p
BC149 BC157	10p	BF196 BF197	10p 12p	2N3704 2N3705	10p
BC158	10p	BF200	28p	ZN3706	10p
BC159	10p 12p	BFR39 BFR79	24p	2N3708	10p
BC182 BC183	12p	BFX29	26p 22p	2N3710 2N3819	10p 28p
BC184	120	BFX48	32p	2N3904	15p
BC212 BC213	14p	BFX84 BFX88	22p 22p	2N3906 2N6027	15p 55p
BC214	140	BFÝ50	18p	2N6028	60p
BC441 BC461	32p	BFY51 BFY52	18p	40673	65p
BC547	10p	BRY39	40p		
74 Serie	S TTL				-
7400 7401	12p 14p	7447 7450	84p	7410. 74121	37p 36p
7402	14p	7451	14p	74122	51p
7404	17p 23p	7453 7454	14p 14p	74123 74132	64p 56p
7406	28p	7460	14p	74141	63p
7408 7410	14p	7472 7473	29p 29p	74150 74151	173p
7410	28p	7474	29p	74151	79p 144p
7414	62p	7475	51p	74155	73p
7420	14p 36p	7476 7483	29p 91p	74157 74159	66p 200p
7430	14p	7485	132p	74164	126p
7432	28p 36p	7486 7490	40p 46p	74174 74179	110p 120p
7438	36p	7491	75p	74180	120p
7440 7442	15p	7492 7493	52p	74190 74191	188p
7445	65p 88p	7493	52p 73p	74191	158p 120p
7446	88p	7496	85p	74193	120p
C-M				74367	120p
4000	18p 18p	4018	84p 90p	4054 4055	100p 110p
4002	18p	4023	18p	4080	96p
4007	18p	4024	64p	4071	18p
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ı	AC126	21p	BC547	14p	MJ2955	99p	BY127	16p	CA3011	86p	4015
i	AC127	21p	BC548	14p	MJE340	64p	OA47	9p	CA3014	190p	4016
ı	AC128	21p	BC549	14p	MJ E 29 5 5	99p	OA70	9p	CA3018	86p	4017
ı	AC176	21p	BC557	14p	MJE3055	68p	OA79	14p	CA3020	174p	4018
ı	AC187 AC188	23p 23p	BCY34 BCY70	74p 17p	OC28 OC35	108p	0A81	15p	CA3023	186p	4019
ı	AD149	65p	BCY71	20p	OC71	19p	OA85 OA91	14p 7p	CA3028 CA3030	85p 120p	4020
l	AD161	50p	BCY72	17p	OC84	46p	OA95	9p	CA3035	190p	4022
1	AD162	50p	BD115	59p	TIP29	50p	0A200	9p	CA3036	115p	4023
ı	AF114	30p	BD124	97p	TIP30	52p	OA202	9p	CA3042	198p	4024
ı	AF115	27p	BD131	47p	TIP31	59p	IN914	4p	CA3045	275p	4025
ı	AF116	25p	BD132	47p 43p	TIP32 TIP33	59p	IN916	6р	CA3046	90p	4026
ı	AF117 AF118	23p 30p	BD135 BD136	50p	TIP34	60p	IN4001 IN4002	6p	CA3048	205p 154p	4027
ı	AF124	32p	BD137	50p	TIP35	253p	IN4003	6p	CA3049 CA3052	152p	4029
ı	AF125	32p	BD138	51p	TIP36	389p	IN4004	7p	CA3054	73p	4030
ľ	AF126	32p	BD139	51p	TIP41	59p	IN4005	10p	CA3078	136p	4031
ľ	AF139	40p	BD140	52p	TIP42	59p	IN4006	10p	CA3080	57p	4032
ı	AF239	45p	BF173	30p	TIP2955	126p	IN4007	11p	CA3081	148p	4033
ı	BC107 BC108	9p 9p	BF179 BF180	37p 37p	TIP3055 ZTX108	64p 12p	3A-50V	16p	CA3086	47p	4035
ł	BC109	9p	BF181	37p	ZTX300	12p	3A-100V 3A-200V	16p 18p	CA3090 CA3130E	329p 113p	4041
ı	BC147	10p	BF182	37p	ZTX500	12p	3A-400V	20p	CA3140E	74p	4043
ı	BC148	10p	BF183	37p	2N1305	25p	3A-600V	25p	CA3160E	140p	4044
	BC149	10p	BF194	13p	2N1306	39p	10A-50V	25p	LM301	39p	4045
ı	BC157	11p	BF195	13p	2N1711	21p	10A-100V	25p	LM308	54p	4046
	BC158	11p	BF196 BF197	13p 16p	2N2219 2N2904	21p 26p	10A-200V	28p	LM380	106p	4047
	8C159 8C168	11p 12p	BF200	36p	2N2905	26p	10A-400V 10A-600V	36p 42p	LM381	120p	4049
ı	BC169	12p	BF224	16p	2N2906	20p	10A-600V	42p	LM710 LM741	41p 24p	4050 4054
	BC171	12p	BF257	37p	2N2926G	9p	BRIDG	GE	LM747	73p	4055
	BC172	12p	BF258	40p	2N2926R	9p	REC	S	LM748	36p	4056
	BC173	12p	BF259	44p	2N2955	99p	1A 50V	22p	LM3900	73p	4060
	BC177	16p	BFR39	30p	2N3053	17p	1A 100V	24p	MC1304	292p	4066
	BC178 BC179	16p	BFX29 BFX30	34p 38p	2N3054 2N3055	54p	1A 200V	27p	MC1310	197p	4067
	BC182L	13p	BFX84	27p	2N3702	10p	1A 400V	30p	MC1458	78p	4068
	BC183L	11p	BFX85	29p	2N703	10p	1A 600V	34p 34p	MC1496 NE555	33p	40 69
	BC184L	13p	BFX86	31p	2N3704	12p	2A 50V 2A 100V	36p	NE556	76p	4071
	BC212L	13p	BFX87	31p	2N3705	12p	2A 200V	38p	NE561	385p	4072
	BC213L	13p	BFX88	29p	2N3706	12p	2A 400V	40p	NE562	385p	4076
	BC214L	14p	BFY50 BFY51	17p 17p	2N3708 2N3709	12p 11p	2A 600V	42p	NE565	138p	4077
	BC238 BC303	14p 30p	BSX20	21p	2N37U9 2N3711	10p	TRIA	20	NE566	126p	4078
	BC328	15p	BU105	191p	TIS43	40p	2A 100V	26p	NE567 SN76003	175p	4081
	BC338	15p	BU208	299p	2N2646	56p	2A 100V	42p	SN76013	185p	4086
			D THE	LICTO			2A 400V	71p	SN76023	160p	4501
			R. THY				6A 100V	50p	SN76033	259p	4502
	1A/50V	28p	5A/400V	59p	10A/400V		6A 200V	66p	TAA621	257p	4507
	1A/200V	38p	5A/600V	75p	10A/600V		6A 400V	81p	TAA661	175p	4510
	1A/400V	40p	5A/800V 7A/100V	82p 59p	10A/800V		10A 100V 10A 200V	68p	TBA540 TBA641	236p 281p	4516 4 5 18
	1A/600V 3A/100V	58p 36p	7A/200V	68p	16A/200V		10A 200V	113p	TBA800	102p	4519
	3A/200V	38p	7A/400V	120p	16A/400V		10A 400V		TBA810	113p	4532
	3A/400V	51p	7A/600V	123p	16A/600V			130p	TBA820	90p	4539

	Section 1	-					_
	C 84	O.C.	TT	L. 13p	DI	L. SOC	VETC
ı	CM	03	7400	12-			KEIS
1	4000	16p	7400	13p	8 pin	10p	16 pin
	4001	18p			14 pin	13p	28 pin
	4002	18p	7402	16p		OPTO	
	4006	110p	7403	17p	TH 200		
	4007	20p	7404	22p 22p	TIL209	Red	
	4008	97p	7405	22p	TIL209	Yellow	
	4009	62p	7406	44p	TIL209	Green	
	4010	62p	7407	44p	-2	Red	
	4010	40-	7408	22p	-2	Yellow	
	4011	19p		23p	-2	Green	
	4012	20p	7409	23p	TIL209 Clip		
ī	4013	59p	7410	16p	·2 Clip		
3	4014	108p	7411 7412	24p	· Z Clip		_
	4015	97p	7412	24p		NEON	S
	4016	59p	7413	38p	Red		
	4017	99p	7414	66p	Green		
	4018	99p	7416	33p	Amber		
	4019	65p	7417	33p	Neon Bulb		
	4020	119p	7420	18p			
	4021	99p	7425 7427	28p	R	ESISTO	DRS
	4022		7423	28p	0.3 watt 10	O to 1 MO	F24
		94p	7430		0.5 watt 10	O to 10M	D E 24
	4023	21p		20p	2.5 watt 0.	220 44 10	O weller
	4024	83p	7432	28p	2.5 Watt O.	221110110	11 00/00
	4025	20p	7437	29p	5 watt 4.70		
	4026	190p	7441	74p	10 watt 10	T to TUKM	w/w
	4027	59p	7442	74p			
	4028	96p	7445	123p 123p		NTION	
	4029	118p	7446	123p	1ΚΩ-2ΜΩ 9	S/gang Lin-	-Log
	4030	62p	7447	102p	5KΩ-2MΩ S	S/gang Sw	itch Loa
	4031	249p	7448	130p	5ΚΩ-2ΜΩΙ	Dong Lin	-Loa
		108p	7454	19p			
	4032					SLIDER	35
	4033	157p	7470	39p	5KΩ-500K	OS/G Lin-I	Log
	4035	130p	7472	28p	5KΩ-500KI	D/G Lin-	Loa
	4041	87p	7473	39p	Presets sub	min 100	0-5M0
	4042	85p	7474	33p	Slider Bezze		
	4043	99p	7475	48p	Slider knobs		
	4044	96p	7476	40p			
	4045	150p	7480	46p	CA	PACIT	ORS
	4046	131p	7482	91p	Min cerami		
	4047	95p	7483	90p	22 pf to 33		4p
	4049	65p	7485	106p	ZZ pi 10 00	oo pi	
	4050	oop	7486	36p	C	WITCH	IFS
		65p		30p		SPS	
	4054	120p	7489	399p	Large		
	4055	140p	7490	57p 82p	Toggle	SPD) T
	4056	140p	7491	82p	0.00	DPD	71
	4060	117p	7492	57p	STD	SPS SPD	
	4066	70p	7493	57p	Toggle	SPD	T
	4067	410p	7494	83p		DPC	
	4068	24p	7495	70p	Min	SPS	
	4069	22p	7496	83p	Toggle	SPD	T
	4070	60p	74100	130p	00	DPC	
	4071	22p	74107	36p	C. off	DPC	T
	4072	23p	74118	125p	Sub. min.	SPS	T
		167-		200	Toggle	SPC	T
	4076	167p	74121	38p	, oggie	DPC	
	4077	65p	74122	55p	Death button	. CRC	길
	4078	24p 23p	74122 74123 74144	103p	Push butto	SPS SPC	T
	4081	23p	74144	345p		SPL	7
	4082	23p	74145	83p		DPD	71
	4086	24p	74153	62p	Bias push n		
		19p	74154	138p	Bias push b	rake	
	4501		74165	115p	Stide sub.m	in. DPD	
		130p					
	4502	130p	74100	132n	Slide std.	DPE	T
	4502 4507	130p 59p	74174	115p 132p 79p	Slide Std. Slide C. off		
	4502 4507 4510	130p 59p 132p	74174 74175	79p	Slide Std. Slide C. off	DPE	
	4502 4507 4510 4516	130p 59p 132p 132p	74174 74175 74180	79p 85p	Slide C. off	DPI	T
	4502 4507 4510 4516 4518	130p 59p 132p 132p 114p	74174 74175 74180 74181	79p 85p 295p	Slide C. off	Y (Adjus	stable s
	4502 4507 4510 4516 4518 4519	130p 59p 132p 132p 114p 60p	74174 74175 74180 74181 74192	79p 85p 295p 131p	ROTAR 1P/2-12 w	Y (Adjus	stable s
	4502 4507 4510 4516 4518	130p 59p 132p 132p 114p	74174 74175 74180 74181	79p 85p 295p 131p 131p	Slide C. off	Y (Adjus	stable s

D.1.	L. SUCI	CEID	44-	LUUD	PEARE	H2
	10p	6 pin	14p	1-in 1	3Ω	72
14 pin	13p 2	28 pin	30p	2 3 111	Ω	72p
	OPTO					725
				2in	Ω	72p
TIL209	Red		14p		ΒΩ	64
TIL209	Yellow		24p	2 lin 1	ΒΩ	64
TIL209	Green		24p	21in 64	4Ω	72
-2	Red		18p	24in 1	3Ω	76
-2	Yellow		30p	21in 80	Ω	118
.2	Green		30p			
TIL209 Clip	Oleen		4p	TRANS	FORME	RS
-2 Clip			5p	Mains p	rim 220-24	0v
	NEONS	3		000	100M/A	02.
Red			39p	6-0-6		93
Green			39p	9-0-9	75M/A	97
				12-0-12	100M/A	99
Amber			39p	0-6-0-6	280M/A	150
Neon Bulb			20p	0-12-0-12	150M/A	150
D	ESISTO	DC		6-0-6	1.5A	345
			- 0	9-0-9	1A	280
0.3 watt 10	(1 to 1 ML)	E24	2p	9-0-9	2A	398
0.5 watt 10 2.5 watt 0.	Q to 10ML	E24	2+p		1A	317
2-5 watt 0-	22Ω to 10Ω	2 w/w	23p	12-0-12		
5 watt 4.70	to 10KQ v	v/w	21p 23p 12p	15-0-15	1A	350
10 watt 10) to 10KQ	w/w	15p	20-0-20	2A	544
10 11011 101	1 10 1011			25-0-25	2A	585
DOTE	NTIOM	ETER	2	30-0-30	2A	625
				12V	2A	367
1ΚΩ-2ΜΩ 9	o/gang Lin-	Log	27p			
5KΩ-2MΩ S	gang Swi	tch Log	55p	Antex X2 5		385
5KΩ-2MΩ I	D/gang Lin-	Log	70p	Antes C2 1	watt iron	
	CLIDED			Antex C2 1	watt Iron	385
	SLIDER	3		Antex CX 1		385
5KΩ-500K			73p	Antex stand		155
5KΩ-500KI	D/G Lin-L	.og	88p	Desoldering	gun	640
Presets sub.	min. 1000	2-5MΩ	10p	Spare nozzi	e	90
Slider Bezze			21p	PP3 Batt cli	ins	6
Slider knobs			10p	PP9 Batt cli	ne	10
				Ferric chlori	do	95
CA	PACIT	DRS		Faint Chion	de	81
Min cerami				Etching pen		
22 pf to 33		40	each	Etching kit		245
ZZ pi 10 00	оо рі		00011	Telephone p		70
S	WITCH	FS		Ear piece st		16
	SPST		27p	Ear piece cr	ystal	33
Large			29p	Speaker clo	th, black	
Toggle	SPD		Zab	36in x 2		120
	DPD	T	35p	001172	OIII)	
STD	SPST		66p			
Toggle	SPD	T	70p	0 -	-	
	DPD	T	75p	I-OR	AD W	46
Min	SPST	г	54p		AP. K	
Toggle	SPD		58p			711
roggic	DPD		70p	224425		MARK
0 -4					D NEW DE	
C. off	DPD		90p	A pack of	useful gene	ral pu
Sub. min. Toggle	SPS1		48p	pose comp	onents.	
Toggle	SPD'		52p	Resistors 1	O off noch	
	DPD	Т	60p	resistors	0 011 69011	140
Push button	n SPS	r	60p	Value	12 10Ω-8-2 100 mixe	IVILI
	SPD		69p	Capacitors	100 mixe	d troi
	DPD		91p	100 pf to 1	000µF	
Bias push n			15p	5xNPN		n. pur
			20p	10xIN914		
Bias push b	DAD.	T	13	2×741	1×555	
Stide sub.m	in. DPD	-	13p	3xsockets	L.E.D's	
Slide std.	DPD		13p	Switches	Batt clins	
Slide C. off	DPD	T	15p	Audio TRY	Batt, clips Tuning cap	
-				Releve	Esta Cap	
ROTAR	Y (Adjus		(qor	Relays	Etc.	
1P/2-12 w	av 2P/	2-6 way		EXCE	LLENT VAL	UE

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125p post and packingl.

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KITS FOR SYNTHESISERS, SOUND EFFECTS



COMPONENTS SETS include all necessary resistors. capacitors, semiconductors, potentiometers and transformers. Hardware such as cases, sockets, knobs, keyboards, etc. are not included but most of these may be bought separately. Fuller details of kits, PCBs and parts are shown in our lists.

CIRCUIT AND LAYOUT DIAGRAMS are supplied free with all PCBs unless "as published

PHOTOCOPIES of all P.E. texts for most of the kits

PHONOSON

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P.E. MINISONIC Mk. 2 SYNTHESISER
A portable mains-operated Miniature Sound Synthesiser, with keyboard circuits. Although having slightly fewer facilities than the large P.E. Synthesiser the functions offered by this design give it great scope and versatility. Consists of 2 log VCOs. VCP. 2 envelope shapers. 2 voltage controlled amps, keyboard hold and control circuits. HF oscillator and detector, ring modulator, noise generator.

mixer, power supply.
Set of basic component kits Set of printed circuit boards

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The well acclaimed and highly versatile large-scale mains-operated Sound Synthesiser complete with keyboard circuits. Other circuits in our fists may be used with the Synthesiser to good advantage.

The Main Synthesiser: PSU. 2 linear VCOs, 2 ramp generators, 2 input amps, sample hold, noise generator, reverb amp, ring modulator, peak level circuit, envelope shaper, voltage controlled amp.

Set of basic component kits Set of printed circuit boards

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GUITAR EFFECTS PEDAL (P.E. July 75)
Modulates the attack, decay and filter characteristics of an audio signal not only from a guitar but from any audio source, producing 8 different switchable effects that can be further modified by manual controls. Possibly the most interesting of all the low-priced sound effects units in our range. Circuit does not duplicate effects from the Guitar range. Circuit Overdrive Unit.

Component set with special foot operated switches Alternative component set with panel switches Printed circult board

SOUND BENDER (P.E. May 74)

A multi-purpose sound controller, the functions of which include envelope shaper, tremolo, voice-operated fader, automatic fader and frequency-doubler.

Component set for above functions (excl. SWs)

£7.84

Printed circuit board Printed circuit board
Optional extra—additional Audio Modulator, the use of
which, in conjunction with the above component set, can
produce "jungle-drum" rhythms.
Component set (incl. PCB)
£2-88

PHASING UNIT (P.E. Sept. 73)

A simple but effective manually controlled unit for introducing the "phasing" sound into live or recorded Component set (incl. PCB)

PHASING CONTROL UNIT (P.E. Oct. 74)
For use with the above Phasing Unit to automatically control For use with the above Phasin the rate of phasing. Component set (Incl. PCB)

SOPHISTICATED PHASING AND VIBRATO UNIT

A slightly modified version of the circuit published in Elektor'. December 1976, and includes manual and ultiomatic control over the rate of phasing and vibrato. Component set 177-69 Printed circuit board £2.33

WAH-WAH UNIT (P.E. Apr. 76)
The Wah-Wah effect produced by this unit can be controlled manually or by the integral automatic controller.
Component set (incl. PCB)

POST AND HANDLING

Automatically produces Wah-pedal and Swell-pedal sounds each time a new note is played.

Component set. PCB, special foot switches

Component set and PCB, with panel switches

£4.83

PUSI AND MANDLING
U.K. orders—under £15 add 25p plus VAT, over £15 add 50p
plus VAT. Keyboards £2.00 plus VAT.
Optional Insurance for compensation against loss or damage
in post, add extra 50p for cover up to £50, £1.00 for £100 cover,

£2.00 for £200 cover.

Elre, C.I., B.F.P.O., and other countries are subject to

P.E. JOANNA PLUS ORGAN VOICING

The basic five octave electronic plano (P.E. May/Sept 75 and Sound Design) has switchable alternative voicings for Honky-Tonk, ordinary plano, and Harpsichord or a mixture of any of these three, together with facilities including fast and slow tremolo loud and soft pedal switching, and sustain pedal switching. The modification retains all the circuitry associated with the piano but in addition provides an organ-voice envelope facility with 5 switchable pitches, variable attack and sustain, phasing and vibrato.

Set of components lexcl switches for PSU, Frequency generator, Pitch and Note Divider, Envelope Shapers, Volcings, and Control circultries. (Order as KIT 71-5) £109-75 Set of PCBs (Order as PCB SET 71-6)

SYNTHESISER TUNING INDICATOR (P.E. July 77)
A simple 4-octave frequency comparator for use with synthesisers and other instruments where the full versatility of the P.E. Tuning Fork is not required.

Component and PCB (but excl sw.)

GUITAR FREQUENCY DOUBLER (P.E. Aug. 77) A modified and extended version of the circuit publishing Component set and PCB

GUITAR SUSTAIN (P.E. Oct 77)

Maintains the natural attack whilst extending note duration Component set, PCB and foot switches £4-96 Component set, PCB and panel switches

WIND AND RAIN UNIT

manually controlled unit for producing the above-named mponent set (Incl. PCB)

GUITAR OVERDRIVE UNIT (P.E. Aug. 76)
Sophisticated, versatile Fuzz unit, including variable and switchable controls affecting the fuzz quality whilsts retaining the attack and decay, and also providing filtering. Does not duplicate the effects from the Guitar Effects Pedal and can be used with it and with other electronic Instruments.
Component set using dual slider pot £6:80
Component set using dual rotary pot £6:20
Printed circuit board £1:62

FUZZ UNIT

Simple Fuzz unit based upon P.E. "Sound Design" circult Component set (incl. PCB)

TREMOLO UNIT

Based upon P.E. "Sound Design" circuit.
Component set (incl. PCB)

TREBLE BOOST UNIT (P.E. Apr. 76)
Gives a much shriller quality to audio signals fed through it.
The depth of boost is manually adjustable.
Component set (incl. PCB)

£2.40

DON'T FORGET VATI

apply to export orders).

P.E. TUNING FORK (P.E. Nov. 75)
Produces 84 switch-selected frequency-accurate tones. A
LED monitor clearly displays all beat note adjustments: Ideal
for tuning acoustic or electronic musical instruments.
Main component set (incl. PCB)
Power supply set (incl. PCB)
£7.03

SEE OTHER PAGE FOR KEYBOARDS, AND OUR LISTS FOR OTHER COMPONENTS AND ACCESSORIES STOCKED

P.E. SYNCHRONOME (P.E. Mar. 76)

An accented-beat electronic metronome, providing duple, triple and quadruple times with full control over the beat rate. Can also be used as a simple drum-beat rhythm generator. Includes power supply.

Component set (incl. loudspeaker)

Printed circuit board

£2.04

TAPE NOISE LIMITER

TAPE NOISE LIMITER
Very effective circuit for reducing the hiss found in most tape recordings. All kits include PCBs
Standard tolerance set of components
Superior tolerance set of components
Regulated power supply (will drive 2 sets)
£4-69

ENVELOPE SHAPER WITHOUT VCA (P.E. Oct. 75)

Provides full manual control over attack, decay, sustain and release functions, and is for use with an existing voltage controlled amplifier.

Component set (incl. PCB)

ENVELOPE SHAPER WITH VCA (P.E. Apr. 76)
This unit has its own voltage controlled amplifler and has full manual control over attack, decay, sustain and release functions.

Component set (incl. PCB)

TRANSIENT GENERATOR (P.E. Apr. 77)

An envelope shaper, without VCA, having the usual attack, decay, sustain and release functions, and in addition it also provides a "Repeat Effect" enabling a synthesiser to be programmed to imitate such instruments as a mandolin or banjo

Component set
Printed circuit board

Table 182

WAVEFORM CONVERTER

Slightly modified from a circuit published in "Elektor". Converts a saw-tooth waveform into four different waveforms: sine-wave, mark-space saw-tooth, regular triangle form, and squarewave with an externally variable mark-space ratio.

Component set (incl. PCB but excl. sw/s)

£8-19

VOLTAGE CONTROLLED FILTER (P.E. Dec. 74)
Part of the P.E. Minisonic now released as an independent kit for use with other synthesisers.

Component set (incl. PCB) (Order as Kit 65-1)

RING MODULATOR (P.E. Jan. 75)
Part of the P.E. Minisonic now released as an independent kit for use with other synthesises.
Component set (Incl. PCB) (Order as Kit 59-1)

NOISE GENERATOR (P.E. Jan. 75)

Part of the P.E. Minisonic now released as an independent kit for use with other synthesisers.

Component set (incl. PCB.) (Order as Kit 60-1) £3-35

SOPHISTICATED POWER SUPPLIES

A wide range of highly stabilised low noise power supply kits is available—details in our lists.

MICROPHONE PRE-AMP (P.E. Apr. 77)
Component set (incl. PCB)

£3.78

VOICE OPERATED FADER (P.E. Dec. 73)

For automatically reducing music volume during talk-over—particularly useful for Disco work or for home-movie shows. Component set (incl. PCB)

DYNAMIC RANGE LIMITER (P.E. Apr. 77)
Automatically controls sound output to within a preset

level.
Component set (Incl. PCB)

EXPORT ORDERS are welcome, though we advise that a current copy of our list should be obtained before

ordering as it also shows Export postage rates. All payments must be cash-with-order, in Sterling and preferably by International Money Order or through an English Bank, To obtain list send 50p.

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Add 121% (or current rate if changed) to full

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AND OTHER PROJECTS

PHOTOGRAPHS in this advertisement PHOTOGRAPHS in this advertisement show two of our units containing some of the P.E. projects built from our kits and PCBs. The cases were built by ourselves and are not for sale, though a small selection of other cases is available.

LIST—Send stamped addressed envelope with all U.K. requests for free list giving fuller details of PCBs, kits and other components.

OVERSEAS enquiries for list: Europe send 20p: other countries—send 50p.



KIMBER-ALLEN **KEYBOARDS AND CONTACTS**

Kimber-Allen Keyboards as required for many published circuits. The manufacturers claim that these are the finest moulded plastic keyboards available. All octaves are C to C, the keys are plastic surjen-loaded fitted with actuators and mounted on a cohust alturible frame.

3 Octave (37 notes)	£25.50
Octave (49 notes)	£32-25
Octave (61 notes)	£39.75

Contact Assemblies (gold-clad wire) for use with the above keyboards (1 required for each

Type GJ: Single-pole change-over	each 24p
Type GB: 2 pairs of contacts, each pair normally open	each 27p
Type GC: 3 pairs of contacts, each pair normally open	each 36p
Type GE: 4 pairs of contacts, each pair normally open	each 45p
Type GH: 5 pairs of contacts, each pair normally open	each 57p
Type 4PS: 3 pairs of contacts plus single-pole changeover	each 53p

Printed Circuit Boards for use with GJ, GB and 4PS contacts (thus eliminating much interwiring) are available. Details in our lists.

RHYTHM GENERATOR

15-Rhythm Tempo, Timing and Logic control unit (e	xcl. sw's
but incl. PCB)	£12.90
10-Instrument Effects circuits	£13.56
PCB for Effects circuits	€4.25
Power Supply Incl. PCB	£12-00

128-NOTE TUNE-PROGRAMMABLE SEQUENCER

(P.E. Nov/Dec 77)

Enables a voltage controlled synthesiser to automatically play pre-programmed tunes of up to 32 pitches and 128 notes long. Programs are keyboard initiated and note length and rhythmic pattern are externally variable. (Please use order codes quoted in

Main Circuit (Nov) excl. sw's (KIT 76-1)	€20.60
Power Supply (KIT 76-3)	€6.05
Trigger Inverter and Alt. Output (KIT 76-2)	£1.35
LED Counter (KIT 76-4)	£2.45
PCB (as published) for KITS 76-1 & 3 (PCB 76A)	£2-61
PCB for KITS 76-2 & 4 (PCB 76B)	£2.54

P.E. STRING-ENSEMBLE (P.E. commencing Mar 78)

The new keyboard string-instrument synthesiser.	
Power Supply Basic component set	£9.22
Tone Generators (incl. Test components)	£14.93
PCB for PSU and Tone Generator	£3.40
Details of further kits and PCBs in our list.	

FORMANT SYNTHESISER (Elektor 1977/78)

Very sophisticated music synthesiser for the advanced con-structor who puts performance before price. Details in our lists.

3-CHANNEL SDUND-TO-LIGHT (P.E. Apr. 76)

A simple but effective sound-to-light controller capable of operating 3 lamps each of approximately 700 watts. Includes power supply, thyristors, and by-pass switches. Component set (incl. PCB)

DISCOSTROBE (P.E. Nov. 76)

4-channel light-show controller giving a choice of sequential, random, or full strobe mode of operation. Basic component set Printed circuit board

BIDLDGICAL AMPLIFIER (P.E. Jan./Feb. 73)

Multi-function circuits that, with the use of other external equipment, can serve as lie-detector, alphaphone, cardiophone

Pre-Amp Module Components set (incl. PCB)	£4.22
Basic Output Circuits—combined component set	t
with PCBs, for alphaphone, cardiophone, frequency	1
meter and visual feed-back lampdriver circuits.	£6.59
Audio Amplifier Module Type PC7	£7.75

10% DISCOUNT VOUCHER (PE67)

TERMS: Correctly costed, C.W.O., U.K. orders over £40 goods value. Valld until end of month on cover of P.E. This voucher must accompany order.

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TRANSISTORS

AC128

26p

AL 1/0	ZOD
BC107	14n
BC108	145
BC109	14p
BC109C	15p
BC177	14n
BC177	120
BC187	250
BC204	14p
BC209C	140
BC213	150
BC209C BC213 BC214	150
	350
BC262	
BC415 BC478	16p
BC478	29p
BD131	44p
BD132	
BF244A	24p
BF245A	24p
BSY95A	22p
BSY95A1	72 p
OC71	200
OC72	SUP
RPY58A	48p
TIS43	
ZTX108	9p
ZTX301	13n
7TY394 1	610
ZTX3841 ZTX5012N2219	440
Z1X501	13p
2N2219	27p
2N2646	50p
2N2905	35p
2N2905 2N2905A	360
2N2906	235
2N2907	
2N3054	ьбр
2N3055	60p
2N3702	120
2N3704	
2N3819	350
2113013	CAP
2N3820	04p
2N3823E	
2N5459	45p

INTEGRATED CIRCUITS

301 8-pin DIL	48p
318 8-pin DIL	230p
320-15	58p
324 14-pin DIL	87p
341-15	195p
709 8-pin DIL	48p
723 TO5	105p
723 14-pin DIL	56p
726 TO5	980p
741 8-pin DIL	32p
748 8-pin DIL	63p
4024 14-pin DIL	461p
4069 14-pin DIL	18p
4136 14-pin DIL	126p
7805 TO220	205p
7806 TO220	205p
7808 TO220	205p
7812 TO220	205p
7815 TO220	205p
7818 TO220	205p
AY10212 16-pin DIL	650p
AY16721/6	195p
CA3046 14-pin DIL	90p
CA3080 8-pin DIL	82p
CA3084 14-pin DIL	209p
M252 16-pin DIL	680p
MC3340 8-pin DIL	150p
MCM6810 24-pin DI	LB/Up
SG3402N 14-pin DII	- Z0ZD
STK025	595p
TDA 1022 16-pin DIL XR2207 14-pin DIL	672p 420p
ZN425E 16-pin DIL	375p
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BY-2628
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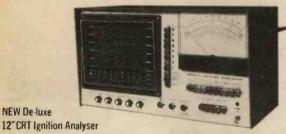
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7404	0.13	7496	0.60	74186 7.20		74LS85	0.90	74LS365	0.50	4060	0.98
7405	0.13	7497	2.38	74188 2.70		74LS86	0.35	74LS366	0.50	4066	0.48
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	0.14	74107	0.28	74193 1.0		74LS109	0.36	74LS670	2.00	4070	0.17
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7411	0.18	74110	0.46	74195 0.84		74LS113	0.36		- 44	4072	0.17
7412	0.21	74111	0.70	74196 0.9		74LS114	0.36	4000	0.14	4073	0.17
7413	0.25	74116	1.60	74197 0.9		74LS123	0.82	4001	0.15	4075	0.17
7414	0.54	74118	0.82	74198 1.4		74LS124	2.45	4002	0.16	4076	1.05
7416	0.27	74119	1.30	74199 1.4		74LS125	0.44	4006	0.92	4077	0.46
7417	0.27	74120	0.82	74221 1.5		74LS126	0.44	4007	0.16	4078	0.22
7420	0.13	74121	0.25	74273 2.1		74LS132	0.69	4008	0.92	4081	0.17
7421	0.28	74122	0.40	74279 1.2		74LS136	0.40	4009	0.45	4082	0.20
7422	0.17	74123	0.53	74283 1.7		74LS138 74LS139	0.53	4010	0.48	4085	0.72
7423	0.25	74125	0.44	74284 6.8		74LS139	0.53	4011	0.15	4086	0.76
7425	0.20	74126	0.45	74293 1.3		74LS151	1.05	4012	0.16	4089	1.55
7426	0,25	74128	0.62	74298 1.9		74LS153	0.50	4013	0.42	4093	0.65
7427	0.25	74132	0.68	74390 1.9		74LS154	1.20	4014	0.80	4094	1.80
7428	0.34	74135	0.68	74393 2.1	Z	74LS155	0.86	4015	0.77	4095	1.10
7430	0.13	74136	0.75	_		74LS156	0.86	4016	0.42	4096	1.10
7432	0.24	74137	0.94			74LS157	0.47	4017	0.77	4097	3.50
7433	0.32	74141	0.58	74LS00 0.1		74LS158	0.53	4018	0.87	4098	1.12
7437	0.24	74142	2.00	74LS01 0.1		74LS160	1.22	4019	0.42	4099	1.90
7438	0.24	74143	2.00	74LS02 0.1		74LS161	0.69	4020	0.92	4404	1.00
7440	0.13	74144	2.00	74LS03 0.1		74LS162	1.22	4021	0.82	4412	0.30
7441	0.52	74145	0.64	74LS04 0.2		74LS163	0.69	4022	0.82	4428	0.80
7442	0.55	74147	1.30	74LS05 0.2		74LS164	1.20	4023	0.15	4445	1.50
7443	0.90	74148	1.18	74LS08 0.1		74LS168	2.00	4024	0.66	4449	0.30
7444	0.90	74150	0.99	74LS09 0.1		74LS169	2.00	4025	0.15	4501	0.17
7445	0.70	74151	0.60	74LS10 0.1		74LS170	1.76	4026	1.28	4502	0.88
7446	0.70	74153	0.60	74LS11 0.1		74LS173	1.05	4027	0.50	4507	0.50
7447A	0.64	74154	1.05	74LS12 0.1		74LS174	1.12	4028	0.67	4508	2.25
7448	0.60	74155	0.63	74LS13 0.4	6	74LS175	1.05	4029	0.86	4510	1.05
7450	0.13	74156	0.63	74LS14 1.1		74LS189	2.85	4030	0.48	4511	0.98
7451	0.13	74157	0.63	74LS15 0.1	9	74LS190	0.81	4031	2.34	4512	0.92
7453	0.13	74159	1.70	74LS20 0.1		74LS191	0.81		1.25	4514	2.85
7454	0.13	74160	0.80	74LS21 0.1		74LS192	1.80	4034	2.00	4515	2.80
7460	0.13	74161	0.80	74LS22 0.1		74LS193	1.80	4035	1.00	4516	1.02
7470	0.28	74162	0.80	74LS26 0.2		74LS195	1.12	4036	2.40	4518	0.99
7472	0.22	74163	0.80	74LS27 0.4		74LS196	1.20	4037	0.99	4519	0.50
7473	0.26	74164	0.89	74LS30 0.1		74LS197	1.20	4038	1.00	4520	1.05
7474	0.26	74165	0.89	74LS32 0.2		74LS221	1.12	4039	2.80	4521	2.00
7475	0.30	75166	0.99	74LS37 0.2		74LS247	0.97	4040	0.88	4522 4527	1.35
7476	0.26	74167	2.70	74LS38 0.2		74LS248	0.97	4041	0.77		1.60
7480	0.45	74170	1.68	74LS40 0.1		74LS249	0.97	4042	0.72	4528	0.92
7481	0.90	74172	4.00	74LS42 0.5		74LS251	1.00			4529	
7482	0.80	74173	1.18	74LS47 0.9		74LS253	1.05	4044	0.82	4536	3.56
7483	0.72	74174	0.89	74LS48 0.9		74LS257	1.05	4045	1.40	4553	4.20
7484	0.90	74175	0.68	74LS49 0.9		74LS258	1.05	4046	1.32	4555	0.85
7485	0.88	74176	0.88	74LS51 0.1		74LS266	0.39	4047	0.96	4556	0.85
7486	0.26	74177	0.88	74LS54 0.1		74LS273	0.50	4048	0.60	4558 4566	
7489	2.00	74178	1.20	74LS55 0.2		74LS279		4049			0.75
7490	0.35	74179	1.10	74LS73 0.3		74LS283	1.00	4050	0.42	4583	
7491	0.65	74180	0.90	74LS74 0.3	4	74LS289	2.85	4051	0.84	4585	1.03

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WATTS!

HY5 Preamplifier

The HY5 is a mono hybrid amplifier ideally suited for all applications. All common input functions (mag Cartridge, tuner, etc.) are catered for internally, the desired function is achieved either by a multi-way switch or direct connection to the appropiate pins. The internal volume and tone circuits merely require connecting to external potentiometers (not included). The HY5 is compatible with all I.L.P power amplifiers and power supplies. To ease construction and mounting a P.C. connector is supplied with each pre-amplifler.

FEATURES: complete pre-amplifier in single pack: multi-function equalisation: low noise: low

distortion; high overload; two simply combined for stereo.

APPLICATIONS; hi-fl; mixers; disco; gultar and organ; public address

APPLICATIONS: ni-ni, mixers, cisco, guitar and organ, public acoress.
SPECIFICATION: Inputs-magnetic pick-up 3mv: ceramic pick-up 3mv: tuner 100mv. microphone 10mV; auxiliary 3-100mV; niput Impedance 47kΩ at 1kHz. Outputs—tape 100mV; main output 500mv.

N.M.S. Active Tone Controls—treble ± 12dB at 10kHz: bass ±12dB at 10kHz. Distortion—0-1% at 1kHz, signal/noise ratio 58dB. Overload—38dB on magnetic pick-up. Supply voltage—±16-50v. Price £5-22 + 65p VAT. P. & P. free

HY5 mounting board B.1. 48p + 6p VAT. P. & P. free

HY30 15W into 8Ω

The HY30 is an exciting New kit from I.L.P. It features a virtually indestructible I.C. with short circuit and thermal protection. The kit consists of: I.C., heatsink, P.C. board, 4 resistors, 6 capacitors, mounting kit, together with easy to follow construction and operating instructions. This amplifier is ideally suited to the beginner in audio who wishes to use the most up to date technology available.

FEATURES: complete kit: low distortion: short, open and thermal protection: easy to build APPLICATIONS: updating audio equipment, guitar practice amplifier, test amplifier; audio oscillator. SPECIFICATION: Output Power—15W R.M.S. into 8Ω. Distortion—0-1% at 15W. Input Sensitivity—500mV. Frequency Response—10Hz-16kHz - 3dB.

Price £5.22 + 65p VAT. P. & P. free

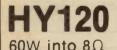
HY50 25W into 8Ω

The HY50 leads I.L.P.'s total integration approach to power amplifier design The amplifier features an integral heatsink together with the simplicity of no external components. During the past three years the amplifier has been refined to the extent that it must be one of the most reliable and robust High Fidelity modules in the World. FEATURES: low distortion: integral heatsink: only five connections. 7 amp output transistors: no

external components
APPLICATIONS: medium power hi-fi systems, low power disco, guitar amplifier

SPECIFICATION: Input Sensitivity—500mV Output Power-25W R. N. S. into 8Ω Load Impedance—4-16Ω. Distortion—0.04% at 25W at 1kHz. Signal Noise Ratio—75dB. Frequency Response—10Hz-45kHz - 3dB. Supply Voltage—±25V. Size—105 × 50 × 25mm.

Price £6-82 + 85p VAT. P. & P. free



The HY120 is the baby of I.L.P.'s new high power range, designed to meet the most exacting requirements including load line and thermal protection this amplifier sets a new standard in modular design

FEATURES: very low distortion, integral heatsink, load line protection, thermal protection, five connections, no external components.

APPLICATIONS: hi-fi: high quality disco, public address, monitor amplifier, guitar and organ.

SPECIFICATION: Input Sensitivity—500mV Output Power—60W R.M.S. into 8Ω. Load Impedance—4-16Ω. Distortion—0.04% at 60W at 1kHz. Signal Nolse Ratio—90dB. Frequency Response—10Hz-45kHz -3dB. Supply Voltage—±35V. Size—114 × 50 × 85mm

Price £15.84 + £1.27 VAT. P. & P. free



The HY200 (now improved to give an output of 120 watts) has been designed to stand the most rugged conditions such as disco or group while still retaining true hi-fi performance. FEATURES: thermal shutdown: very low distortion, load line protection; integral heatsink; no external

Text UNES; thermal shutown, very low assistance to a components.

APPLICATIONS: hi-fit disco: monitor, power slave: Industrial, public address

SPECIFICATION: Input Sensitivity—500mV. Output Power—120W R.M.S. Into 8Ω. Load Impedance—4-16Ω. Distortion—0.05% at 100W at 1kHz. Signal/Noise Ratio—96dB. Frequency Response—10Hz-45kHz-3dB. Supply Voltage—±45V. Size—114 × 50 × 85mm.

Price £23-32 + £1-87 VAT. P. & P. free

HY400 240W into 4Ω

The HY400 is I.L.P.'s 'Big Daddy' of the range producing 240W into 4Ω' It has been designed for high power disco or public address applications. If the amplifier is to be used at continuous high power levels a cooling fan is recommended. The amplifier includes all the qualities of the rest of the family to lead the market as a true high power hi-fidelity power module.

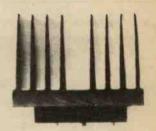
FEATURES: thermal shutdown, very low distortion; load line protection, no external components. APPLICATIONS: public address, disco. power slave; industrial SPECIFICATION: Output Power—240W R.M.S. into 4Ω. Load Impedance—4-16Ω. Distortion—0-1% at 240W at 1kHz. Signal/Noise Ratio—94dB. Frequency Response—10Hz-45kHz - 3dB. Supply Voltage — ± 45V. Input Sensitivity—500mV. Size—114 × 100 × 85mm.

Price £32.17 + £2.57 VAT. P. & P. free

POWER SUPPLIES: PSU36—suitable for two HY30s £5-22 + 65p VAT. P. & P. free. PSU50—suitable for two HY50s £6-82 + 85p VAT. P. & P. free. PSU90—suitable for two HY120s £13-75 + 1-10 VAT. P. & P. free. PSU90—suitable for one HY200 £12-65 + £1-01 VAT. P. & P. free. PSU180—suitable for two HY200s or one HY400 £23-10 + £1-85 VAT. P. & P. free. I.L.P. Electronics Ltd., Crossland House, Nackington, Canterbury, Kent CT4 7AD









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Description OPDT miniature slide DPDT standard slide Toggle switch SPST	No. 1973 1974		Price £0.11° £0.14°
Toggle switch DPDT	1975		€0.33•
Tamp 250V a.c. Rotary on-off mains switch Push switch – Push to make Push switch – Push to break	1976 1977 1978 1979		£0.42° £0.50° £0.13° £0.18°
ROCKER SWITCH A range of rocker switches SPST – moulded in high insulation. Material available in a choice of colours ideal for small apparatus.	Colour RED BLACK WHITE BLUE YELLOW LUMINOUS	No. 1980 1981 1982 1983 1984 1985	Price £0.30° £0.30° £0.30° £0.30°
Description Miniature SPST toggle, 2 amp	No.		Price
250V a.c. Miniature SPST toggle, 2 amp	1958		£0.50°
250V a.c. Miniature DPDT toggle, 2 amp	1959		£0.55°
250V a.c. Miniature DPDT toggle, ceritre	1960		£0.70*
off, 2 amp 250V a.c. Push button SPST, 2 amp	1961		£0.85°
250V a.c. Push button SPST, 2 amp	1962		£0.78*
Push button DPDT, 2 amp	1963		£0.83.
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Description 20mm x 5mm chassis mounting

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Description 1 pole 12 way 2 pole 6 way 3 pole 4 way 4 pole 3 way	Order No. 1965 1966 1967 1968	Price £0.48° £0.48° £0.48° £0.48°
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QUICK B	LOW 20mm					
Type 150mA 250mA 550mA 800mA	No. 611 6p 612 5p 613 5p 614 7p	1A 1.5A 2A	No. 615 5p 616 6p 617 5p 618 6p	Type No 3A 619 4A 620 5A 62	5p 6p	
ANTI-SU	RGE 20mm					
Type 100mA 250mA 500mA	No. 622 623 624	Type 1A 2A 1.6A All 7p	No. 625 626 627 each	Type 2-5A 3-15A 5A	No. 628 629 630	
QUICK B	LOW 1ain					
Type 250mA	No. 631	Type 500mA All 7p		Type 800mA	No. 634	
Type 1A 2A	No. 635 637	Type 2.5A 3A	No. 638 639	Type 5A	No. 642	
All 6p each						

CASES AND BOXES

INSTRUMENT CASES. In two sections vinyl covered top and sides, aluminium bottom, front and back.

No.	Length	Width	Height	Price
155	8in	5-lin	2in	£1.25°
156	11in	6in	3in	£2-12°
157	6in	4 lin	1 gin	£1-30°
158	9in	4 in 5 in	2 in	£1.76°

ALUMINIUM BOXES. Made from bright ali., folded construction each box complete with half inch deep lid

and str	Evvo.			
No.	Longth	Width	Height	Price
159	5 in	2 lin	1 in	62p°
160	4in	4in	1 in	62p°
161	4in	2 lin	1 in	62p°
162	5 in	4in	1-lin	70p°
163	4in	2 in	2ín	64p°
164	3in	2ín	1in	44p°
165	7in	5in	2 kin	£1.04°
166	8in	6in	2½in 3in	£1.32*
167	6in	4in	2in	86p°

MIDGET WAFER SWITCHES

1965 - 1 pole 12 way	48p°
1966 - 2 pole 6 way	48p°
1967 - 3 pole 4 way	48p°
1968 - 4 pole 3 way	48p°

TRANSFORMERS

MINIATURE MAINS Primary 240V

No.	Secondar		Price
2021	6V-0-6V		90p°
2022	9V-0-9V	/ 100mA	90p°
2023	121/0-12	V 100mA	95p°
2020	12,4-0-12	¥ 10011114	35p
TOTAL ST			
	URE MAINS Primary		
with two	independent secondary	windings	
No.	Туре		Price
2024	MT280-0-6V.0-6	SV DAGC	£1.50°
	MT150-0-12V. 0	121/BLIC	
2025	M1150-0-12V. U	- IZV RIVIS	£1.50°
1 AMPN	AAINS Primary 240V		
No.	Secondary	Price	
2026	6V-0-6V 1 amp	£2.50*	P. & P. 45p
2027	9V-0-9V 1 amp	£2 00°	P. & P. 45p
2028	12V-0-12V 1 amp	£2-60°	P. & P. 55p
2029	15V-0-15V 1 amp	£2.75°	P. & P. 66p
2030	30V-0 30V 1 amp	£3 45°	P. & P. 86p
2000	oot o oot i amp	20-40	1. d 1. oop

STANDARD MAINS Primary 240V Multi-tapped secondary mains transformers available in $\frac{1}{2}$ amp, 1 amp and 2 amp current rating. Secondary taps are 0-19-25-33-40-50V.

Voltages available by use of taps: 4, 7, 8, 10, 14, 15, 17, 19, 25, 31, 33, 40, 25, 0-25v.

No.	Rating	Price	
2031	amp	£5.50*	P. & P 86p
2032	amp	£6.60°	P. & P. 86p
2033	2 amp	£8.40°	P. & P. £1-10
		20.10	

AUDIO LEADS

	MODIO ELMBO	
107	FM Indoor Ribbon Aerial	£0.60°
113	3.5mm Jack plug to 3.5mm jack plug.	
	Length 1.5m	£0-75°
114	5 pin DIN plug to 3.5mm. Jack connected	
	to pins 3&5. Length 1.5m	£0.85°
115	5 pin DIN plug to 3.5mm. Jack connected	
	to pins 1 & 4. Length 1.5m	£0.85°
116	Car aerial extension. Screened insulated	
2	lead. Fitted plug & skt.	£1.10*
117	AC mains connecting lead for cassette	100000
	recorders & radios. 2 metres	£0.68°
118	5 pin DIN phono plug to stereo	
	headphone jack socket	£1.05°
119	2+2 pin DIN plugs to stereo jack socket	
	with attenuation network for stereo	
	headphones. Length 0.2m	£0 90.
120	Car stereo connector. Variable geometry	1711
	plug to fit most car cassette. 8 track	
	cartridge & combination units. Supplied	
123	with inline fused power lead and instructions.	£0.60°
123	6.6m Coiled Guitar Lead Mono Jack Plug	
124	to Mono Jack Plug BLACK	£1.50°
124	3 pin DIN plug to 3 pin DIN plug.	80 754
125	Length 1.5m	€0.75*
123	5 pin DIN plug to 5 pin DIN plug.	€0.75*
126	Length 1.5m 5 pin DIN plug to Tinned open end.	F0.12.
120	Length 1.5m	€0.75*
127	5 pin Din plug to 4 Phono Plugs.	E0.73
14.	All colour coded, Length 1.5m	£1.30°
128	5 pin DIN plug to 5 pin DIN socket.	21.00
.20	Length 1.5m	£0.80°
129	5 pin DIN plug to 5 pin DIN plug mirror	
	image, Length 1.5m	£1.05*
130	2 pin DIN plug to 2 pin DIN Inline socket.	
1.00	Length 5m	€0.68*
131	5 pin DIN plug to 3 pin DIN plug, 1&4	
	and 3&5. Length 1.5m	£0.83°
132	2 pin DIN plug to 2 pin DIN socket.	
	Length 10m	£0.98°
133	5 pin DIN plug to 2 phono plugs.	
	Connected pins 3&5, Length 1.5m	£0.75°
134	5 pin DIN plug to 2 phono sockets.	
	Connected pins 3&5. Length 23cm	£0 68°
135	5 pin DIN socket to 2 phono plugs.	
	Connected pins 3&5. Length 23cm	£0.68*
136	Coiled stereo headphone extension lead.	
	Black. Length 6m	£1.75*
178	AC mains lead for calculators etc.	£0.45°
2000		-

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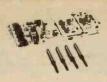
OBA 2BA

High quality audio modules for Stereo and mono

S450

STEREO FM TUNER Fitted with phase lock-loop

+ 40p p&p + 121% VAT



FREQUENCY RANGE	88-108 Mhz
SENSITIVITY	3.0 µV
BANDWIDTH	250 kMz
SPURIOUS REJECTION	50 dB
SELECTIVITY ± 400 kHz	55 dB
AUDIO OUTPUT (22:5 kHz deviation)	100 mV
STEREO SEPARATION	30 dB
SUPPLY REQUIREMENTS	20 to 30V (90mA max)
AERIAL IMPEDANCE	75 ohms
DIMENSIONS	240mm × 110mm × 32mm

The 450 Tuner provides instant programme selection at the touch of a button ensuring accurate tuning of 4 pre-selected stations, any of which may be altered as often as you choose, simply by changing the settings of the pre-set controls. Features include FET Input stage. Vari-Cap diode tuning. Switched AFC LED Sirveo Indicator.

Stereo 30 COMPLETE AUDIO CHASSIS



OUTPUT POWER	7 Watts RMS
LOAD IMPEDANCE	8 ohms
TOTAL HARMONIC DISTORTION	Less than '5% (Typically '3%)
FREQUENCY RESPONSE	50 Hz to 20 kHz ± 3dBs
TONE CONTROL RANGE	± 12 dBs at 100Hz and 10kHz
SENSITIVITY	190 mV for full output
INPUT IMPEDANCE	1 M ohms
TRANSFORMER REQUIREMENTS	22 V.A.C. rated at 1A
DIMENSIONS (Less controls and panel)	200mm × 130mm × 33mm

The Stereo 30 comprises a complete stereo pre-amplifier, power amplifiers and power supply. This, with only the addition of a transformer or overwind will produce a high quality audio unit suitable for use with a wide range of inputs i.e. high quality ceramic pick-up, stereo tuner, stereo tape deck etc. Simple to Install, capable of producing really first class results, this unit is supplied with full instructions, black front panel, knobs, main switch, fuse and fuse holder and universal mounting brackets.

AL60	25	2	25w
UDIO	100	2	R.M.S.
MPLIFIER 10DULE	13	16	
5 Watts RMS E4 ·55 + 2!			
12;% VAT	p p&p	4.3	

OUTPUT POWER	25 Watts RMS
SUPPLY	30-50 V
LOAD IMPEDANCE	8-16 ohms .
TOTAL HARMONIC DISTORTION	Less than . 1% (Typically . 065
FREQUENCY RESPONSE	20 Hz to 30 kHz × 2 dBs
SENSITIVITY	280 mV for full output
MAX. HEAT SINK TEMPERATURE	90°C
DIMENSIONS	103mm × 64mm × 15mm

This high quality audio amplifier module is for use in audio equipment and stereo amplifiers and provides output powers up to 25 RMS with distortion levels below 0.1%.

AL80

AUDIO AMPLIFIER MODULE £7 ·15* + 25p p&p + 8% VAT



125W R.M.S.

OUTPUT POWER	35 Watts RMS
SUPPLY	40-60 V
LOAD IMPEDANCE	8-16 ohms
TOTAL HARMONIC DISTORTION	Less than . 1% (Typically . 06
FREQUENCY RESPONSE	20 Hz to 30 kHz × 2 dBs
SENSITIVITY	280 mV for full output
MAX. HEAT SINK TEMPERATURE	90°C
DIMENSIONS	103mm × 64mm × 15mm

The AL80 is similar in design to the AL60 above and is of the same high quality but provides output powers up to 35W with distortion levels below 0.1%.

AL250 POWER AMPLIFIER



£17 ·25*

10w

OUTPUT POWER 125 Watts RMS continuous OPERATING VOLTAGE 50-80 V LOADS 4-16 ohms FREQUENCY RESPONSE SENSITIVITY FOR 100 WATTS 450 mV INPUT IMPEDANCE TOTAL HARMONIC DISTORTION
50 WATTS into 4 ohms
50 WATTS into 8 ohms

This unit, designated AL250, is a power amplifier providing an output of up to 125W RMS, into a 4 ohm load.

AL30A AUDIO AMPLIFIER MODULES





POWER OUTPUT for 2% THD	10 Watts RMS
TOTAL HARMONIC DISTORTION	Less than ·25 %
LOAD IMPEDANCE	8-16 ohms
INPUT IMPEDANCE	100 K ohms
FREQUENCY RESPONSE	50 Hz-25 kHz ± 3 dBs
SENSITIVITY	75 mV for full output
DIMENSIONS	74mm × 63mm × 28mm

SPM80 STABILISED POWER SUPPLY

£4 .25 + 25pp&p



INPUT A.C. VOLTAGE	33-40V
OUTPUT D.C. VOLTAGE	33 V nominal
OUTPUT CURRENT	10 mA-1·5 amps
OVERLOAD CURRENT	1·7 amps approx.
DIMENSIONS	105mm × 63mm × 30mm

Within ± 1 dB from 20 Hz to 20 kHz ± 15 dBs at 75 Hz

+ 10-20 dBs at 15 kHz Better than 65 dBs (All inputs)

Better than 26 dBs (All inputs)

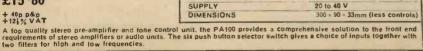
Designed to power two AL60s at 15 Watts per channel simultaneously. Circuit Techniques include full short circuit protection. 20 Hz to 20 kHz × 1 dB -Less than ·1% (Typically ·07%) 100 mV/100 K ohms 100 mV/100 K ohms 3 5 mV/50 K ohms 250 mV

FREQUENCY RESPONSE
TOTAL HARMONIC DISTORTION

PA100



£15 ·80



FOUALISATION

INPUT OVERLOAD

BASS CONTROL RANGE TREBLE CONTROL RANGE SIGNAL/NOISE RATIO

MPA30

MAGNETIC CARTRIDGE PRE-AMPLIFIER



Enjoy the quality of a magnetic cartridge with your existing ceramic equipment using the MPA 30 which is a high quality preamplifier enabling magnetic cartridges to be used where facilities exist for the use of ceramic cartridges only.

3.5 mV for 100 mV output

SENSITIVITY	3.2 ms tot 100 ms ontbut
EQUALISATION	Within ± 1 dB from 20 Hz to 20 kHz
INPUT IMPEDANCE	50 K ohms
SUPPLY	18 to 30 V-re earth
DIMENSIONS	110 × 50 × 25mm (inc DIN socket)

PA12



The PA12 Stereo PreAmplifier chassis is designed and recommended for use with the AL 20/30 A udio Amplifier Modules, the PS12 power supply and the TS98 Transformer. Features include on/off volume, Balance, Bass and Treble controls. Complete with tape output.

FREQUENCY RESPONSE 20 Hz-20 kHz (-3dB) ± 12 dB at 60 Hz BASS CONTROL TREBLE CONTROL 14 dB at 10 kHz 1 Meg. ohm INPUT IMPEDANCE INPUT SENSITIVITY CROSSTALK -60 dB OVERLOAD FACTOR ± 20 dB TAPE OUTPUT IMPEDANCE DIMENSIONS 152mm × 84mm × 25mm

PS12 POWER SUPPLY

Designed for use with the AL30A S.450 and MPA30 in conjunction with transformer T538.

INPUT VOLTAGE OUTPUT VOLTAGE OUTPUT CURRENT

17-20v AC 27-30v DC 800mA 60mm × 43mm × 26mm

£1 .30

GE 100 NINE CHANNEL MONO-GRAPHIC EQUALIZER

The GE100 has nine 1 octave adjustments using integrated circuit active filters. Boost and Cut limits are ± 12dB. Max. Voltage handling 2 V RMS, T.H.D., 0.05%, input impedence 100K. Output impedence less than 10 K. Frequency response 20 Hz - 20 kHz (3dB). The nine gain controls are centred at 50, 100, 200, 400, 800, 1,600, 3,200, 6,400 and 12,800 Hz. The suggested gain controls are 10 K LIN sliders (not suggested gain controls are 10 K LIN sl

SG30 POWER SUPPLY BOARD for GE100 15-0-15 VOLT £5-50 + 12}% VAT + 25p p&p

SIREN ALARM MODULE

American Police screamer powered from any 12 volt supply into 4 or 8 ohm speaker. Ideal for car burglar alarm, freezer breakdown and other security purposes. Order No. S15. No. BP124. Only £3-50 + 8% VAT + 25p p&p

MA60 HI-FI AMPLIFIER KIT

Build you own top quality amplifier, save yourself pounds. The MA60 kit comprises the following Bl-kits modules, 2 × AL60 amps 1 × PA100 pre-amp, 1 × SPM80 stab, power supply, 1 × BMT80 transf. giving 17 watts RMS per channel STEREO. All modules covered by the Bl-PAK satisfaction or money back guarantee. Details of the above modules are in this ad.

TC60 KIT

A beautifully designed genuine TEAK WOOD veneered cabinet to put the professional touches to your home built amplifler. Full set of parts Incl. Front & Back Panels, Knobs, Chassis, Fuses, Sockels, Noen, etc. Ideal for the MA60. Size: 425mm × 290mm × 95mm. Price £19.95 + 121% VAT + 86p p&p

TRANSFORMERS

T538 For use with S.450 AL30A MPA30 Order No. 2036 Price: £3·26 + 55p pap + 12½% VAT Price: £3·26 + 55p pap + 12½% VAT Price: No. 2050 For use with Stereo 30 Order No. 2050 BMT30 For use with AL60 SPM30 Order No. 2034 Price: £5·46 + 86p pap + 12½% VAT BMT250 For use with AL50 Order No. 2034 Price: £6·35 + £1·10 pap + 12½% VAT Order No. 2035



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THE PIONEERS OF MODULAR DISCO/P.A. EQUIPMENT NOW OFFER PACKAGE DEALS AT INCOMPARABLE PRICES

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C/W LIGHT SHOW & DISPLAY. TWIN SPEAKERS & LEADS

Standard 100W

£225 or Deposit £50.80 12 Months @ £19.11 or 24 Months @ £10.66

Super 200W

£275 or Deposit £61.80
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GXL 200W (with twin 200) wort cobinets

£349 or Deposit £77.72 12 Months @ £29.19 or 24 Months @ £16.28

COMPLETE STEREO **ROADSHOWS - BUILT IN** SOUND TO LIGHT/SEQUENCER & DISPLAY

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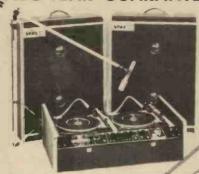


illustration shows GXL Centaur System

These systems feature full mixing for two decks tope & mic with monitoring focilities and ore supplied complete with sound sequencer, disploy, speaker lead exc

JUST PLUG IN AND GO!

BSR Decks - 17,000 Line Loudspeakers - Rugged Aluminium Trimmed Cobines - Lue Light And Phones Output - Slove Output - Deck Lights/Motor Storts (OXL)

MINI DISCO **100 WATT** MONO SYSTEM

£179.50 Deposit £40.66

12 Months @ \$15.45 24 Months @ \$8.61

Similar in appearance to the Centaur and complete with laudspeakers and leads

Headphones to whany system EM007 Elect of Mic £15.00 Boom Stand Co-riage on all disco and Arsystelus (Included in H.P. Prices)

20% Deposit Terms On All Orders Over £150 - 12 or 24 Months - Low Interest

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SA308 30W 8 ohms 45V	£9.95*	SUPPLY FOR £10.90*
SA604 60W 4 ohms 50V	£13,25	TWO MODULES
SA608 60W 8 ohms 65V	£14.25	TWO MODULES SUPPLY FOR
SA1204 120W 4 ohms 75V SA1208 120W 8 ohms 95V	£15.95	ONE MODULE
SA2404 240W 4 ohms 95V	229.50	TWO MODULES SUPPLY FOR ONE MODULE
O 2% Distortion, 30Hz-20, KHz sensitivity 240 mV to suit most exers	2dB, Fully Shart/C	Open Circuit proof input

TOP QUALITY COMPONENTS THROUGHOUT

COMPLETE LIGHTING CONTROL AT YOUR FINGERTIPS!



Lighting Control Unit Mk II 4kW Sequence - Sound Light - Dimmers - Automatic Level Integrated Logic M £44.50 Module £32.50 Panel

Three Channel Sound to Light 3xW 1-240W input - master £26.75 Module 219.75

SPARES & ACCESSORIES - LOUDSPEAKERS & CAB	INETS
Rope Lights - Red or Multicolour £22.00 Melos Echo Chamber	£59.00
per 12 ft. Headphones	£7.50*
Rope Light Controller for up to 120 ft £30.00 Sirens: English Police, USA Pol	
Fuzz Lights-Red/Blue/Yellow/Green £22.80 Destroyer, Alien Voice Simula	
Magnetic Cartridge Equalisers £3.50* Bulgin 8 way lighting plug/so	cket £190

100 Watt Chassis Loudspeakers 12" £23.50 Large 12" £21.50

Empty Loudspeaker Cabinets Small 12" £15.50, Large 2 x 12" £28 Small 2 12" £22.50 1 x 18" £29.50

Projector lamps: A1167 £2.90, M6 £5.65. 100W Spot lamps Red/Blue/Yellow/Green £1.50 ea £13.50 for 10 MD Spot Banks: 3-way 300W £19.50 4-woy 400W £22.50. Bubble machines (optikinetics) £36.50

Strobe tubes 80W £8,50 Kickproof Grille 24" wide £3.25 Metre
Kick Resistant Grill 50" wide £3.25 Metre FULL RANGE OF RE-AN PRODUCTS IN STOCK SEND FOR OUR BROCHURE NOW!!

DISCO MIXERS - COMPLETE OR MODULAR MONO OR STEREO

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connect madule with switch already fitted

FEATURES INCLUDE FEATURES INCLUDE:
Twin Deck - Mic & Tape Inputs Wide rai
& treble controls - Full hemaphone mani
Crossfode - Professional Tanda of ptyto m

£22.50 £33.50 £3.95 £5.50 COMPLETE MIXERS (with case) €39.50 £57.50 £45.75 £63.75 Stereo 18

STROBE UNITS



Bro-Strobe 4-6 Joules £37.50 Super Strobe 2-3 Joules £22.50 (Pro-Strobe has external trigger facility).

PROJECTORS - PLUTO - NEW LOW PRICES!!!

	CHOIC	E OF WHEEL/CASS	ETTE
P150 150W Tungsten	£37.50	Liquid wheels	£7.50
P500 100W Q.I.	£74.95	Cassettes	€8.00
P500 250W Q.I.	£84.95	Picture wheels from	£4.75

(Wide choice available) PIEZO HORNS only £7.50

YES! - only £7.50 (As fitted to our package PA system)

Direct from Motorola Inc., USA at an UNBEATABLE PRICE

No crossover required 4kHz - 30kHz rated 75W/8 ohms 150W/4 ohms use two per 100W amplifier - Full instructions supplied





PACKAGE P.A. SYSTEMS (2 Year Guarantee)

Complete with PIEZO horn columns fitted with 100 watt units (100 watt system-illustrated)

100 Watt £149.50 Deposit £35.26

12 Months @ £12.91 or 24 Months @ £7.20
Includes & Chonnel 100 Watt Amplifier with
Treble, Bass and Master Controls plus Leadand Twin Piezo Horn Columns (shown on right).

200 Watt £225.00

Deposit £50.80 12 Months @ £19.11 or 24 Months @ £10.66 zsix Mixed Inputs plus Three Sets of Bass and Treble Controls plus Slave Output and Moster Control

ACCESSORIES Melos Echo Unil £59.00

A high quality Cassette Tape Echo Unit giving long tape life, infinitely variable echo depth and speed control. Suitable for all mics, and

High quality Boom Stand £15.50 Floor Stond £9.90. ECM81 Condense Mic. – Removoble Leod – Good Anti-Feedbock £19.95.* EM507 Condense Mic. – Good Volue £15.00. Phasers £19.80.



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Make your own mixer - Mono/Stereo - up to 20 channels with these, easy to wire modules - Available as PCB's or assembled on panels.



Input Stages Up to 20	Mono £5.95	Mono C/W £8.95
	Sterea £9.50	Stereo C/W panel etc. £12.50
Mixer/Monitor (One only per system)	Mono £5.95	Mono (/W panel etc. £8.95
por 2/210m/	Stereo £9.50	Stereo C/W £12.50

£9.50

Send for free brochure for complete specification

Saxon AP100 Amplifier £45

Four mixing inputs - 100W into 4 ohms Wide range bass & treble controls master - Twin outputs

Power supply for

up to 20 channels

Saxon 150 Amplifier £59

Four mixing inputs - 100W into 8 ohms 150W into 4 ohms - wide range bass & treble controls + maste



£1.00

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bit mos rams 95p, TBA800 80p, CD4051 45p, 723 14 pin I.C.'s 35p.
DIODES, 8 Y127 9p, IN4002 4p, IN4005 7p. 600V 3 amp 17p, Lucas bridge recs. 400V
1.5 amp 30p.
MAN 3A 3mm led displays 50p, Min. Nixie 5870ST 75p, Pot core unit. has six pot cores including one FX2243 (45mm) and two FX2242 (35mm) 3 TO3 sil. power transistors on heat sink. 3 20mm panel fuseholders and panel with various transistors. diodes and a 5 amp plastic SCR. 51.75 plus 75p postage.
MOTORS, Model type 1.5-6 volts 20p, 'BIG INCH' sub min motor 115v AC, 3 rpm 25p.
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CONDENSER MIKES OMNI, 1K IMP fuses deaf aid battery, supplied) £4.95.
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PROJECT BOXES, BLACK ABS PLASTIC WITH BRASS INSERTS AND LID.
75 x 56 x 35mm 44p, 95 x 71 x 35mm 52p, 115 x 95 x 36 60p.
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BUZZERS, GPO open typp 3-6 v 30p. Large plastic domed type loud note 6 or 12 volts 50p,
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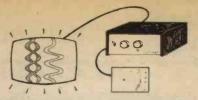
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BC 337	each	.15	Aerial lead	,	.45
28 pin socket	each	.45	Power supply	3	3.25
	each	.25	Complete kit, no extras ne	eded £4	7.50



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TV GAMES

IN THIS issue we present what we feel is the best combination of TV games available. The P.E. TV Game Centre has 14 different game options and can provide hours of amusement for a single player or for the whole family. When we tested this unit we found that a couple of four-year-olds could play at least one of the games and, in fact, they got very upset when we took the Centre away to work on the article. The Centre has also provided an afternoon's enjoyment for a large family and even "grandad" got quite enthusiastic -well he was unable to watch the football!

We must point out that with colour, sound-through-the-set and all those game options, the unit is quite complex and not one for the beginner to tackle. However, there is much enjoyment to be gained from both building and using the unit. (The main part of the project is in this issue and concludes with construction details next month).

TV ENGINEERING

A recent lecture by Dr Boris Townsend, head of the IBA's Engineering Information Service, has given rise to various comments in the technical press about some of his predictions. We have taken the trouble to bring you most of what he said in this issue. We do not feel that the predictions require further comment from us as Dr Townsend is in a much better position to make such predictions than we are, and no doubt most readers will have their own opinions anyway.

To many the review of the present state of the art may be an eye opener but we are sure that to most readers the 1984 and 1989 sections will start some thoughts.

Having made a few predictions in the past and been hounded by them later, we do not intend to start exploring the future of this or other fields. Reference to the editorial of March '78 is recommended for those who would like to try this particular pastime—we will not be drawn into the trap.

Part of Dr Townsend's lecture which, due to space restrictions, we have had to omit said:

"Samuel Butler once said that a hen is merely an egg's way of making another egg. Now, I don't know how many of you would regard that as a valid statement, but I warn you that I

may not say anything which has greater validity—nor indeed shall I say much that is capable of proof, though perhaps some of my remarks may turn out to have been correct.

Neither am I intent on arguing that an engineer is merely a television studio's way of making another television studio—although financial controllers firmly believe that studios do spawn engineers. No, any engineer is far too arrogant to denigrate his importance in the scheme of things. But since I am here not so much as an engineer but as a soothsayer, it behoves me to begin modestly, once I have gently indicated that peering into our future may call for a little lateral thinking."

TV COMPETITION

It is also interesting to note that the use of the TV set for purposes other than watching broadcasts will probably mean the TV companies have more to compete with than each other. We could be said to be adding to this competition by presenting our Game Centre—not a bad thing perhaps?

Mike Kenward

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made payable to IPC Magazines Limited.

Letters

Queries regarding articles published in PE should be addressed to the Editor, at the Editorial Offices, and a stamped, addressed envelope enclosed. We cannot undertake to answer questions regarding other items, nor to answer technical queries over the telephone.



This is really two separate t.v. games circuits combined in one box to give a selection of stunt motorcycle games and a large variety of ball games too.

The four "bike" games comprise: Stunt Cycle, Drag Race, Super Stunt Cycle, and Motocross. These are provided by the AY-3-8760-1 games chip, and occupy the right-hand side of the Game Centre fascia.

There are ten ball games comprising Solo Target, Solo Basketball, Solo Squash, Tennis, Double Target, Gridball, Soccer, Ice Hockey, Basketball and Squash.

STUNT CYCLE SECTION

These are games for one player who controls the speed of a motorbike and rider. At the start of each game, the motorbike and rider are stationary at the upper left-hand side of the TV screen. As the player turns the throttle controller, the motorbike and rider move across the screen on track 1. The motorbike sound starts with the bike movement and as the bike and rider accelerate, the motorbike sound reflects these speed changes. The motorbike wheels have an appearance of rotating at a speed also related to the throttle setting. At the end of track 1, the bike and rider reappear on track 2 at the left-hand side, and likewise at the end of track 2, the bike appears on track 3 at the left-hand side of the screen. Movement of the bike and rider on track 3 over the righthand edge of the screen causes a reinitialisation of the bike and rider to the left of the screen on track 1. There is no further movement until the throttle is reset to a slow speed and then increased again. The playing field for each game is shown opposite.





DRAG RACE

The object of this game is to reach the end of track 3 in the shortest time. The three-digit score is automatically reset as the rider first begins to move on track 1 and the score is incremented until the game is over. The score appears centred on the screen above track 1, and remains until the start of the next game.

Drag Race requires a speed shifting to achieve the lowest time scores. As the throttle speed is increased and the rider begins to move, the bike is in speed one and moves at a set rate across the screen. The only way to accelerate the bike motion is to return the throttle to a "slow" position and then return to a "fast" position. This shifting procedure will move the bike into speed 2, which will then go across the screen at a faster rate. Another "shift" will allow speed 3.

A PROFESSIONAL/AMATEUR switch is provided to select a difficulty factor. In the professional mode, a crash occurs if the player tries to increase the throttle speed too rapidly. A crash will flip the bike and rider upside down, and the sound will be a high-pitch screech. At the end of the crash, the bike and rider are reinitialised on track 1, and the score reset. In the easy mode, no crash is allowed.

MOTOCROSS

As the throttle speed is increased, the bike and rider move across track 1 at a rate determined by the throttle controller setting. Motocross has no speed shifting. Located on each of the three tracks are obstacles. The PRO'/AM' option switch selects the number of obstacles per track. The easier mode has one obstacle per track and the harder mode has two obstacles per track.

The object of this game is to traverse the three tracks in the shortest time, doing a "wheelie" over each obstacle. The score counters record the run time in the same manner as the Drag Race game.

In Motorcross the crash is not caused by accelerating too rapidly. The crash is caused by not doing a "wheelie" over an obstacle. In the "wheelie" position, the bike will have the front wheel lifted off the track. A crash into an obstacle will flip the bike upside down and produce the screech sound. The score resets at the end of the crash.

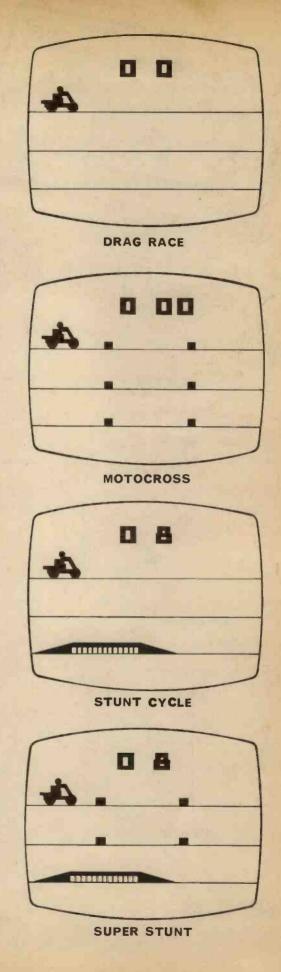
STUNT CYCLE

The object of this game is to control the throttle speed to properly jump the ramp and buses located on track 3. The game begins with 8 buses and with each successful jump over the ramp and buses, an additional bus appears. The game is over when the maximum number of errors has been reached, which is 3 or 7 errors depending on the position of the PRO'/AM' switch or when 36 ouses have been jumped, in which case the screen will fill up with buses. The game is then started by reselecting the Stunt Cycle game input.

Errors are caused by accelerating too rapidly, insufficient speed to clear the buses, or landing too far past the back ramp after the jump. The bike and rider flip upside down and a screeching sound indicates an error. The score records the number of errors in the first digit and the number of displayed buses in the next two digits.

SUPER STUNT CYCLE

This game is similar to Stunt Cycle with the addition of obstacles on track 1 and track 2. The object is to do a "wheelie" over each obstacle and then adjust the throttle for the correct speed to jump the buses on track 3. The PRO'/AM' option switch selects 2 obstacles per track and allows 3 errors per game in the harder mode, and 1 obstacle per track and 7 errors per game in the easy mode. Errors are



caused by accelerating too rapidly, not being in the "wheelie" position over the obstacles, insufficient speed to clear the buses, or landing too far past the back ramp after the jump. The score records the number of errors and the number of buses displayed, the same as in the game of Stunt Cycle.

BALL AND PADDLE SECTION

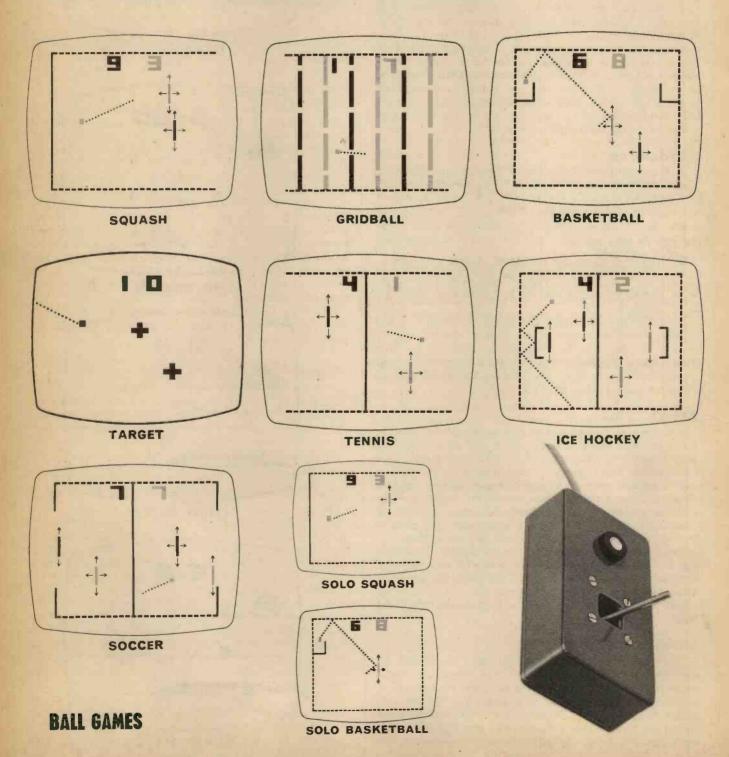
In all games, the ball starts at slow speed. If the high speed mode has been selected the ball will switch to high speed after seven consecutive hits by the players without a goal being scored.

The bats are segmented into five zones, each zone defining

a different rebound angle. The zones listed from top of bat to bottom are nominally 40° up, 20° up, horizontal, 20° down, 40° down. A ball passing through a bat from behind will have its angle influenced as above, but not its left/right direction.

All two player games terminate when one player has 15 points at which time the score flashes and the bats have no further effect on the ball.

A tone of approximately 500Hz, 1kz and 2kHz will be generated for a nominal period of 32ms for "ball hits wall", "ball hits bat" and "score". The output is capable of direct driving a 100Ω speaker.



SQUASH

This game uses the playing area shown opposite. Each player can move over the whole court. The game will start when the player whose service it is, depresses his service button. The ball moves off with a random angle towards the front wall. The colour of the ball will change to the colour code of the next player to hit the ball. Should the wrong player intercept or be hit by the ball it will be considered a fault. Points will only be given if won on player's own service. Points won on opponent's service will only cause a service change.

SOLO SQUASH

This game is for a single player. The right score counts the number of successive hits in the current game (to a maximum of 15), the left score the number of volleys played.

GRIDBALL

This is a game of considerable mental agility. Each player has three sets of vertically moving barriers to block the ball from approaching his end of the field, and openings in the barriers to permit the ball to advance towards the opponent's end. The game starts when both players have depressed their SERVICE buttons. The ball moves away from the face off point with a random angle in either direction.

BASKETBALL

The basketball games use the closed playing area shown. Participant players must deflect the ball and cause it to enter the top of the goal to score. The game starts when both players depress their SERVICE buttons simultaneously. The ball moves from the service point with a random angle in either direction.

SOLO BASKETBALL

Basketball practice is a one player game which utilises only the left basket (see opposite). The right counter displays the number of hits the player makes without scoring while the left counter shows the number of baskets made. Play starts when the right SERVICE button is depressed.

TARGET

In the double game each player has a cross which he can move to any point on the screen. From the edge of the field will come moving targets one at a time, and a player must position his cross over the target and simultaneously push the SERVE button to get a hit. Moving the cross towards the origin of the moving target will give a player the chance to beat his opponent to a score, but will also give him less thinking time in which to react when the target flies out. All scores are shown at the top of the screen.

Only one cross appears for the single player version of this game.

TENNIS

The game uses the playing area shown (opposite). Each player can only move around his side of the court. The game will start when the player whose turn it is to serve, pushes his SERVICE button. The service will automatically change every five points scored. At service the ball will move away from the service point with a random angle but always towards the net.

HOCKEY

Forwards on both sides have freedom to move over the entire playing area. The goal keepers will be locked in the horizontal axis in front of their respective goals, but will move in the vertical axis in the same manner as the forwards.

The game starts when both players have depressed their SERVICE buttons. The ball will move away from the face-off point with a randomly selected angle in either direction.

SOCCER

Motion of the players is as in the hockey game. The game will start when the loser of the previous goal depresses his SERVICE button. The ball will move away from the kick-off point with a randomly selected angle but always towards the goal of the winner of the previous goal.

CIRCUIT DESCRIPTION

No simple description can be given for the extensive internal workings of the two l.s.i. games chips, and most of the remaining electronics fall into two categories: game selection and t.v. interfacing, so that the games unit may be plugged directly into the t.v. aerial socket. See Fig. 1.

Because there are two games chips, a system is necessary to switch from one to the other and this is the function of IC9. The ball and paddle games are selected on IC1 by momentarily connecting certain combinations of Strobe (STR) and select (SEL) lines, but when any of these selection switches (S1-8, S10, S11) are pushed, generating a low input to the IC10(a) NAND gate, the output IC10(a) pin 13 goes high which will reset the D-type flip-flop IC9(a.) By the same mechanism IC9(a) can be set via IC10(b) when a motorcycle game is selected.

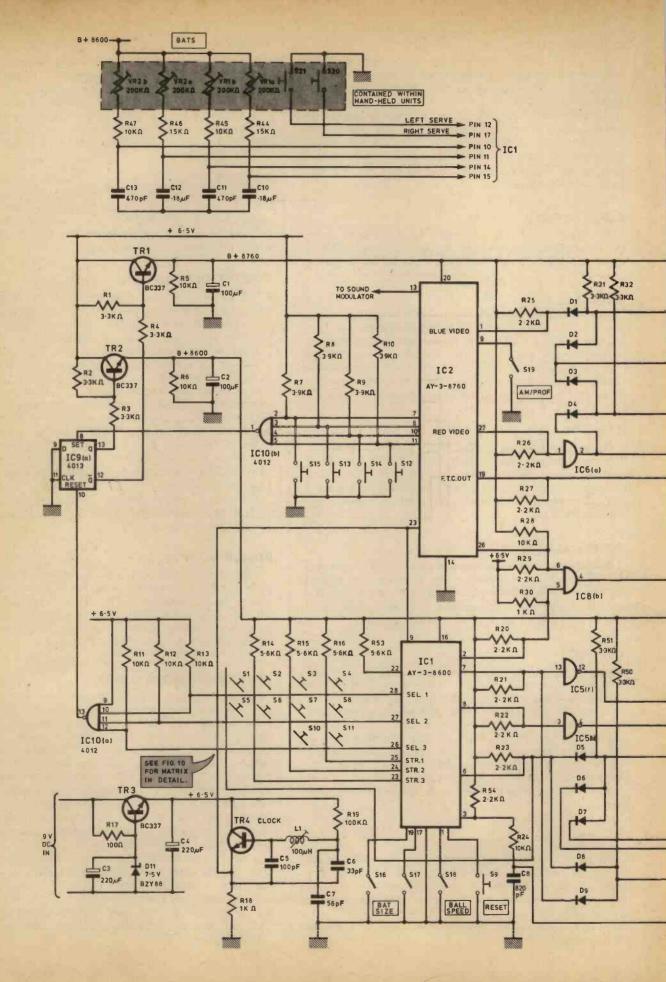
The Q and \overline{Q} outputs from the flip-flop are used to control the 6.5V power supply lines to both games i.c.s, thus shutting down the one not in use.

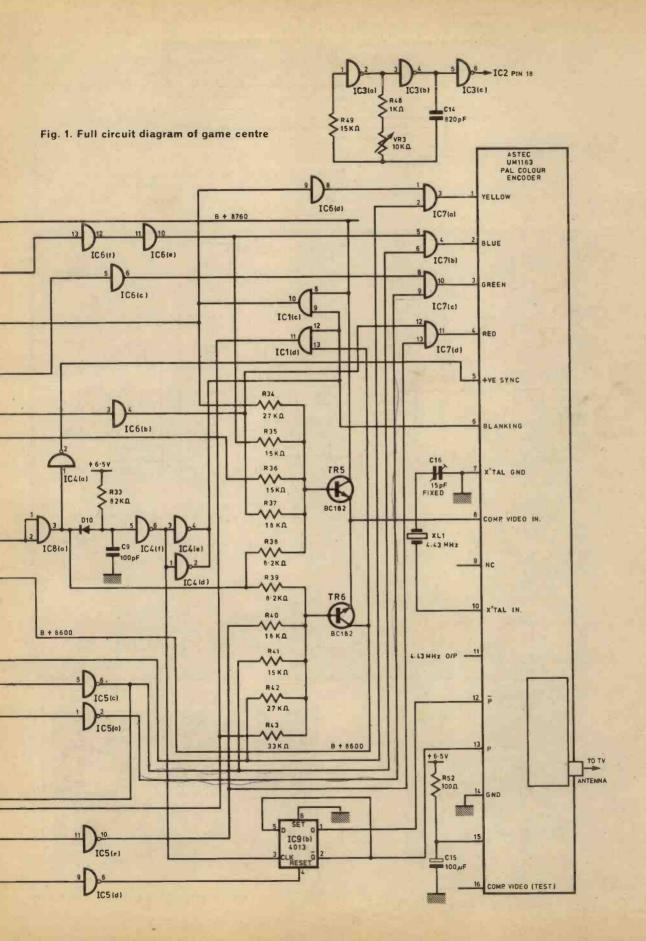
POWER SUPPLY

The d.c. input enters via a miniature jack socket, and is generated by an external mains power pack which plugs directly into a 13A socket. The Zener stabiliser comprising TR3 and D11, etc. produces 6.5V to supply the unit. Both games i.c.s are motivated by a clock signal which is provided by the oscillator designed around TR4.

COLOUR ENCODER UNITS

As can be seen in Fig. 1, the colour encoder has four inputs: yellow, blue, green and red. The bus carrying the video outputs from IC1 can be identified, and these are ORED at IC7 with the video outputs from IC2. This is arranged so that whichever games chip is in use, the colour encoder will receive the necessary signals. It is in this "colour bus" that the colour system departs from a black and white system, where a composite signal only is generated. In this game, the right video output (for right player) is fed to the blue input, and likewise the left video goes to yellow. The players' bats then comply with this colour coding. Some diode or signal processing is also necessary, and can be seen around IC5. A composite signal is generated, and the cycle games composite signal is produced by the summing amplifier TR5. The ball games signals are similarly summated at amplifier TR6. These two composite signals are then ORED by means of the mutual emitters of TR5 and TR6, and fed to the composite video input of the modulator system.





	1 op View		TOP VIEW	L. a
V ₅₅	01	20 Select Input 1	BLACK C 1	28 7 V00
Sync	2	27 Select Input 2	BUPST INTERVAL C. 2	27 WHITE
Blanking	4 3	26 Select Input 3	TEST RSH7 C 3	26 T SYNC
Color Burst	5 4	25 Strobe 1	TEST RSON C 4	25 COLOUR A
Background	ds	24 Strobe 2	TEST REYT C 5	24 COLOUR B
Boundaries .	6	23 Strobe 3	PAL/NISC C 6	33 CLOCK IN
Left Video	₫;	22 Do not connect	M010CR055 🗀 7	22 TEST SCORE
Flight Video		21 E Ball Speed Inhibit	SUPER STUNT C. L. B	SI D HC
3 975MHZ Clock in	9	20 Right Bai Size	PRO/AM OPTION (2 9	20 D NC
Lett Horizontal In	d 10	19 Left Bai Size	ORAG RACE 10	19 FTC OUT
Left Vertical In	ri ii	18 Reset	STUNT CYCLE II	18 THROTTLE
Left Serve		17 Right Serve	POR 📮 12	17 🗆 NC
Sound Output		16 D V _{CC}	SOUND I	16 TEST RSOV
Right Horizontal In		15 Right Varrical in	V _{SS} □ M	15 TEST SOUNO

Fig. 2. Pinouts for games i.c.s (a) AY-3-8600 (b) AY-3-8760

COMPONENTS . . .

The sync and vertical flyback pulses, etc. generated by IC1 and IC2 are combined at AND gate IC8(b) to drive the video modulator directly, and the summing amplifier for inclusion in the composite video signal.

Sound effects come from the television receiver's own loudspeaker, and these signals are modulated onto a 6MHz carrier generated within MOD 1. This r.f. output is fed to the u.h.f. modulator via C17.

UHF VISION MODULATOR

The u.h.f. vision modulator unit MOD 2 receives the sync and video signals from the games chips, and also the r.f. frequency-modulated carrier from the audio modulator MOD 1. The u.h.f. carrier generated within MOD 2 is modulated by these inputs and the carrier is pretuned to the European Channel 36 (591.5MHz).

Resistors	
R1-R4, R31, R32, R50, R51	3-3kΩ (8 off)
R5, R6, R11-R13, R24, R28, R45, R47	10kΩ (9 off)
R7-R10	3.9kΩ (4 off)
R14-R16	5.6kΩ (3 off)
R17, R30, R48, R18	1kΩ (4 off)
R52	100Ω (1 off)
R19	100kΩ (1 off)
R20-R23, R25-27, R29	2.2kΩ (8 off)
R33	8.2kΩ (1 off)
R34, R42	27kΩ (2 off)
R35, R36, R41, R49, R44, R46	15kΩ (6 off)
R37, R40	18kΩ (2 off)
R43	33kΩ (1 off)
All resistors ¼W 5%	

Potentiometers

VR1 200kΩ lin carbon (dual axis) VR2 200k Ω lin carbon (dual axis) VR3 -10kΩ lin carbon

Capacitors

C1, C2 100µF/6·3V (2 off) C3, C4 220µF/10V (2 off) C5 100pF silvered mica (1 off) C6 33pF silvered mica (1 off) 56pF silvered mica (1 off) C8, C14 820pF silvered mica (2 off) 100pF silvered mica (1 off) C9 C10, C12 0.8 µF Mullard C280 (2 off) C11, C13 470pF silvered mica (2 off) C15 100µF/6.3V (1 off) C16 10-40pF trimmer (1 off)

All components are available from Teleplay

Transistors and Diodes

TR1-TR3 BC357 (3 off) TR4-TR6 BC182 (3 off) Silicon signal diode (10 off) D1-D10 BZY88 8V2 (1 off)

Integrated Circuits and Modules

AY-3-8600 (1 off) AY-3-8760-1 (1 off) IC1 1C2 IC3-IC6 CD4069B (hex inverter) (4 off) IC7 CD4071B (quad OR) (1 off) IC8 4081B (1 off) IC9 4013A (dual D/) (1 off) IC10 4012B (quad NAND) (1 off)

Miscellaneous

Module 1 UM1263 sound modulator (Astec) Module 2 UM1163E36 vision modulator (Astec) Module 3 UM1164E36 sub-assembly 1.1 100µH tunable choke (screened)

XL1 4.4336MHz crystal

28-pin i.c. socket (2 off) 14-pin i.c. socket (8 off)

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30mm (2 off)

Plastics box for main unit, 215 × 130 × 70mm sloped (1 off)

Push-to-make switches for S1-S15, S19, S20 (17

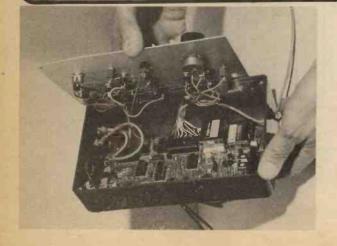
off) Single-pole, single-throw toggle switches for S16-

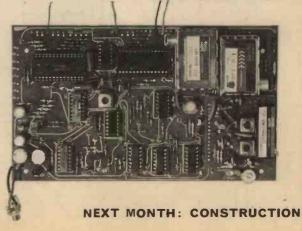
S19 (4 off)

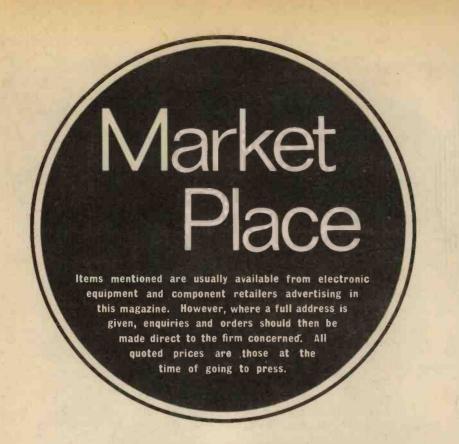
Knob for "throttle" control. Mains power pack (V) Low loss coaxial cable and TV aerial plug

4-way screened cable for hand held units

Printed circuit board







ARALDITE RESIN PACKS

Araldite, the two-part mineral-filled epoxy resin system which is widely used throughout the electronic industry for potting small components, is now available in the form of a two-part sachet containing 200g of material from Ciba-Geigy Plastics.

The pack is in two sections, one containing the resin and the other containing the hardener: the two parts are separated by a clip. When the clip is removed, the two components come into contact with each other and can be mixed thoroughly by manipulation. To use, the bag is pierced at one end and the mixture squeezed out to be used as required.

The performance of the epoxy resins can vary if the two components are not mixed in the correct ratio, and the two-part sachet is a good way of ensuring that the optimum performance is achieved, and as the two components differ in colour the operator has a visual means of checking that the two parts are thoroughly mixed before use.

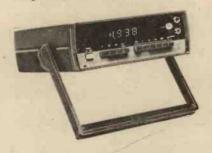
The hardener was chosen to give the system rapid curing properties at moderate temperatures, and is therefore particularly suitable for small castings. The mixture will cure fully at 25°C in between 24 and 36 hours, but will cure in as little as three hours at 60°C. Once mixed, the system has a life at 25°C of about 2-3 hours.

For further information contact Ciba-Geigy Plastics and Additives Company, Plastics Division, Duxford, Cambridge.

DIGITAL MULTIMETER

The new digital multimeter available from Farnell Instruments Ltd will be of interest to anyone who has used a multimeter switched to the incorrect range with disastrous results.

This meter, called the Data Precision 1350 can withstand 500V on its resistance ranges; and a 6,000V spike on any voltage range for 500nS.



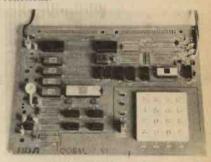
The 1350 offers a basic accuracy of 0·1 per cent and the instrument has a total of 26 ranges; a.c. and d.c. voltages from 100mV f.s.d. to 1,000V, resistance from 100m Ω to 20M Ω and a.c. and d.c. currents from 0·1 μA to 2A.

The instrument is mains powered and has a 0.43 inch 3½ digit l.e.d. display. Recalibration is required once a year with only two adjustments necessary.

The cost of the unit is £115 excluding VAT and further details can be obtained from Farnell International Instruments Ltd, Sandbeck Way, Wetherby, West Yorkshire LS22 4DH.

COMPUTER KIT

A new do-it-yourself computer kit, the COSMAC VIP (Video Interface Processor) has been launched by RCA Limited. The system is designed to interface with a cathode-ray display or, via a suitable modulator, with a TV receiver and allows the user to assemble a complete microcomputer for creating and playing video games, generating computer graphics and developing microprocessor control functions



The VIP offers a complete computer system on a printed-circuit board, using only 4k bits of ROM. Programs can be generated and stored in an audio cassette tape recorder for easy retrieval and use.

At the heart of VIP is RCA's COSMAC CDP1802 microprocessor, incorporating C-mos circuitry for low power consumption and an 8-bit architecture for ease of application.

The CDP1802 microprocessor chip has a 2048 byte RAM, a single-chip graphic video display interface, a built-in hexadecimal keyboard, a 100 byte audio tape cassette interface, power supply and facilities for expanding both memory and input/output interfaces. An interpretive programming language known as CHIP-8 simplifies the programming of video games using hexadecimal code.

The VIP system is readily expandable, both on the printed-circuit board and through connectors. RAM capacity can be doubled on the board to 4096 bytes by adding four 4k bit devices, and can be expanded to 32 kilo bytes by adding further memory capacity through a 44-pin connector socket in the board. Parallel input/output expansion to 19 lines can be achieved for use with music synthesisers, relays, a low-cost printer or an ASCII keyboard.

The VIP hobbyist manual contains detailed information on kit assembly, operating procedures, CHIP-8 interpreter programming technique, machine-language programming, logic description, test programs, troubleshooting guides and system expansion. The manual also includes program listings for 20 video games, with simple instructions to non-programmers for using the hexadecimal keyboard.

VIP is usable with NTSC monochrome TV standards, but a PAL interface chip with colour and programmable sound capability will be available soon.

For further details contact: RCA, Sunbury-on-Thames, Middlesex.

VIDEO AGE BEGINS

Sony's Betamax home video recorder will be in the shops from June at an expected price of £750, with half hour cassettes at £7.

The PAL colour recorder/player can record programmes via its own tuner. In fact it can be timed to record a chosen programme up to three days in advance.

By having its own tuner, recording and viewing of two different channels can take place simultaneously. Also the potential fire hazard of an unattended live TV is avoided.

Recording duration can be set to 15, 30 or 60 minutes, any multiple of 15 minutes, or to the end of the cassette. A 3½ hour cassette will cost £13.50.

Pre-recorded video material will presumably become as readily available as sound cassettes are now. In 1971 Sony and Time Life jointly announced plans to produce video programmes.

With a small video camera and microphone, simple home video programmes can be produced on the Betamax.

During recording a remote pause control enables material to be edited from the armchair.



Sony claim head life of over 1,000 hours with replacement heads very approximately costed at £60. The slanted azimuth head was developed by Sony and the tape

speed is an incredible 19mm per second. Tape width is 13mm. Recorded picture-quality is visibly indistinguishable from transmitted pictures.

TELEPHONE CHARGE CLOCK

A new microprocessor based product will, for the first time, enable telephone users to control the cost of each call as it is being made.

The Monitel telephone charge clock, computes and displays the actual cost of the call in pounds and pence, automatically accounting for the day of the week, the time of day, charge band tariffs and the VAT rate.

The unit is based on the Rockwell PPS4/1 microprocessor range and is programmed by insertion of a punch card. When the telephone rates change a new card is automatically supplied for a nominal fee.

Two models cater for world wide needs;

818

a UK type and an International version. Designed to complement Post Office telephones the Monitel is available in all seven standard colours; ivory, grey, green, black, red, blue and yellow.

It does not infringe the Post Office Act because it is not directly connected to the telephone, but runs off mains power and is activated by the user.

When not being used to monitor calls the Monitel reverts to a 12 hour clock which can be set by keys hidden from view on the top surface of the unit where the telephone is placed.

The only manual selections required by the user are operation of the touch keys on the face of the display which coincide with the Post Office charge bands "I" local, "a" up to 56km, "b" over 56km, which are given in the telephone dialling code booklet, and touching the start/stop keys when the call is connected and again when it is completed.

The International model incorporates six alternatives, the three UK charge bands plus international bands 1, 2 and 4 which cover all of Europe, North America and the Caribbean.

The price of the UK model is £29.00 and the International version £39.00 both inclusive of VAT. Further details are available from Monitel Limited, Berechurch Road, Colchester, Essex.





Practical Electronics July 1978

MULLARD MODULES

Three new economy priced amplifier packages which utilise proven Mullard modules have been introduced by Radio and Television Components Ltd.

Package One consists of two LP1173 audio modules and an LP1182/2 stereo preamplifier module.

Package Two features two LP1173 modules plus an LP1184/2 stereo pre-amp with integral magnetic pre-amp.

Package Three includes two simple LP1173 audio modules LP1182/2 stereo pre-amp LP1179 FM tuner and a 1165 AM/FM IF strip.

Each package comes complete with application sheets detailing the performance and termination of each module. To complete the construction of an amplifier or tuner amplifier, volume controls and power supply can also be obtained from the company.

A very substantial saving over list prices being offered on these units: Package One components are £4.95 (normally £25.50), Package Two £6.95 (normally £27.50) and Package Three £9.95 (normally £37.00). Post and packing is £1.00 for any of the packages.

Further details can be obtained from R & TVC, 323 Edgware Road, London

LINEAR CAPACITANCE METER PROJECT

We are informed that Watford Electronics can supply all the components for the Linear Capacitance Meter project as given in last month's issue. Perplexed hobbyists will be pleased to know that that includes the 1μ F 1% polycarbonate capacitor and the $10M\Omega$ 1% resistor.

Watford Electronics, 35 Cardiff Road, Watford, Herts.

FRENCH VICE

The Nodex Quick Grip can be mounted horizontally on a bench top or on the bench edge. As a vice or a press its patented hand lever locking system makes for quick firm locking or loosening.



V notched jaws enable a wide variety of profiles to be held and the vice is adaptable on all drill supports. Maximum jaw opening is 70mm and jaw area is 70 × 35mm.

At 850gms it is easily carried in a tool kit.

This French vice is only £6.50 and is distributed by Special Products Distributors Ltd., 81 Piccadilly, London WIV OHL.



P.C.B. DRILL STAND

The Mega Photolab is a new, low cost p.c.b. drill stand designed for use with conventionally hand-held p.c.b. drill units. It has been specifically introduced to meet requirements in the production of prototype p.c.b.s.

A strong base of machined cast iron supports precision steel guides by means of which the standard 12V drill is raised and lowered.

Important features are its combined simplicity and accuracy, and the fact that

it will accept Mega and other proprietary drills of 34mm diameter as well as drills such as the Titan, at 41mm.

Additionally, the same basic drill stand is available to special order capable of accepting drills between 20 and 41mm diameter, at marginal extra cost.

The stand will accept p.c.b.s up to 10×9 in.

Priced at £14.50, the PLST-12A is manufactured by Mega Electronics Ltd., 9 Radwinter Road, Saffron Walden, Essex CB11 3HU.

TEN YEARS ON

Bandridge, the component and accessory wholesalers, recently celebrated their tenth anniversary. Operations started in the founder's front room and today the turnover is over £3M. Bandridges "own brand" sales now account for more than half of their turnover.

Most of their customers are one or two man retailers, 90 per cent of the company business being in the UK.

Speed of delivery is thought to be so important that Securicor is used and despatch occurs within 24 hours.

The company's motto is fast and friendly and special effort is made to train staff to be familiar with items, prices and stock availability to ensure quick telephone replies.

Initially they specialised in components only but now accessories are as important. Constructors must be particularly pleased that Bandridge made available the S-DECs, T-DECs, μ -DECs and Blob Boards

New products are shortly to accompany the already established range of audio accessories under the "Bandridge" banner. They include a range of intercoms and a range of quality amplifiers to supplement the already highly successful Quantum 150.

Plenty of literature and a sizeable trade catalogue are available to retailers from Bandridge at 80a Battersea Rise, London, SW11 1EH (01-228 9227).

INSTRUMENT CASES

A new range of instrument cases has just been launched by Vero Electronics.

The new cases are designed to present the same appearance as the existing range, ensuring users of continuity of design, but have been modified to allow the top panel to be easily removed for servicing and calibration.

The range includes 19in and half width models, which are available with or without front handles as required.

The standard colour of the cases is blue/grey but four other colours are available to special order.

For more information about the Mk II D-Case ABO17 contact Vero Electronics Limited, Industrial Estate, Chandler's Ford, Eastleigh, Hampshire.





FRANK W. HYDE

SOVIET ACTIVITY

The launching of the Cosmos research satellite in March marked two important events. The first was that the 1,000th Cosmos was put into orbit with 83° inclination, a period of revolution of 104.9 minutes, apogee of 1,024 kilometres and perigee of 978 kilometres.

This satellite is a navigational vehicle to aid shipping and establish precise parameters and geographic coordinates. Like other satellites of this class Cosmos 1000 has influenced the weather forecasting techniques and aided geological prospecting. Though of the same class the Cosmos series vary considerably in design depending upon the tasks they have to carry out. In most cases the end of life of each vehicle has been to burn up in the higher atmosphere after their individual missions were completed. Some however were designed to have a soft landing return.

The most famous of these was Cosmos 110. This satellite made a 22 day trip with two dogs aboard. The successful return provided vital information for the design of future manned missions. No regular pattern has been followed in the launchings but the general level of activity was increased as the programmes advanced. It took more than ten years for the first 500 of these vehicles to be launched. The second 500 of the programme was accomplished in half that time.

THE SALYUT MISSION

The two astronauts who made the record for the longest weightless mission ceased to feel the effects of the prolonged space conditions a short time after their return to Earth. They celebrated it by going for a walk. This marked the end of

their first re-adjustment to Earth conditions. Some fatigue was felt at first when standing up for a prolonged period. However each day this condition was improved. After four days they began to put on some of the 11lbs they had lost in weight.

Geori Grechko when commenting on his space walk, recalled that the sky had seemed more vast and more colourful than it had appeared from the portholes of the satellite. He also said that the Sun had felt hot in spite of the special space suit. In assessing the work programme both Grechko and Romanenko made reports of the success of the newly designed spacesuits. The load suit and the "Chibis" suit, designed to control the blood flow in the normal pattern on Earth, were highly beneficial. The space "bike" and the running track had also contributed a great deal to their comfort and well-being.

A point was made by the controller of the medical group, Professor Anatoli Yegorov, that the compatability of these two men was a very important factor in the success of the mission and had now contributed enough data to show that man could live quite satisfactorily for over a year, without disturbance of physical or mental conditions.

FUTURE PLANS

A Polish and a German cosmonaut will take part in the next missions, following the Soviet members to start the new series. The exact details are not yet ready for release as further study of the data is still proceeding. It is however certain that another supply ship will be part of the programme and that a regular set of shifts will be the main object of the missions. The fact that efficiency improved with time during the first group of multiple manning, will certainly mean that extra experiments will be added. The timing of the tasks and the days according to the normal Moscow time seemed to contribute not only to the efficiency but also to the high morale. It would appear that the biological clock is also an important factor in success in

Is it perhaps possible that a "special" atmosphere contains some special merit where the efficiency of the biological machine is concerned.

No doubt the new plans will include the dropping of sub-satellites as in the past. Remote control and operation of surveillence, both radar and photographic missions, have covered many fields. All of the missions in the past seem to have had multiple purpose types of programme. Many of the missions have appeared rather puzzling. For example Cosmos 881 and 882 launched by the same rocket and in different conditions returned to earth the same day or shortly after. Similarly Cosmos 997 and 998. Certain of this group of Cosmos vehicles such as 921 and 972 went into an unusual orbit at 75.8° inclination. Cosmos 929 was made to perform extensive changes of parameters. This was

a rather larger vehicle than the others. The success of the geodesy and the weather forecasting enable very close touch to be maintained with the Soviet fleet of ships and submarines in all parts of the world.

It would seem logical that with all the increasing activity the Soviet space programme must include some form of shuttle system. Clearly it would be more economic and it is known that drop tests have been made with a delta-wing spaceplane. One possibility is that, in accordance with their usual testing of unmanned vehicles in assessing new designs, some of the more recent launches are the precursors of a new series of vehicles.

A NEW CANDIDATE

Another candidate for the honour of black hole has now appeared. The "Jet" Galaxy M87 has been under extensive examination at Kitt Peak National Observatory in Arizona; there is an increase in the velocity of matter circling the nucleus. The increase is of the order of 100 kilometres a second. This is the strongest evidence for a massive black hole in the centre of M87.

This work began at Mount Palomar Observatory where J. Kristian and his colleagues found that the galaxy had a brilliant pin point of light.

The next step in obtaining confirmation of this theory will be to scan other elliptical galaxies for similar evidence of massive black holes.

COMETS AND ICE AGES

Comets have been very much in the news of late so it is not surprising that there should be a follow up from Fred Hoyle and Wickramsinghe. They have suggested that passing comets may have been responsible for large masses of dust particles injected into the solar system. These are perhaps the cause of a cooling process. It would be necessary for the Earth to pass through the dense inner halo of the comet. If it is assumed that there is an average of say five comets per year which come within one astronomical unit of the Earth, the chances of a mass acquisition of cometary dust is of the order of 108. Such an order of real happening would agree very roughly with the chronological order of geology.

It could be possible that the size of the particles would lead to an effective reduction in light reaching the Earth. Many things could happen, from the reduction of photosynthesis to the disruption of the food chains within a short time. The picture painted by Hoyle and Wickramsinghe of an initial fast freezing which would see the rapid demise of large animals but the longer continued existence of seeds and nuts with small animals and freshwater organisms surviving. This situation does roughly compare with the conclusions for the past eras. The recovery it is suggested would be quite rapid, something of the order of 1,000 years.



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•

A useful quad bus strip (EXP4B) further

	Model	Length"	Width"	Centre channel"	5-way tie points	Bus	Price All units are 0.330' deep.
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	EXP350	3.6	2.1	0.3	46(230)	2(40)	£4.21 UK Orders.
П	EXP600	6.0	2.4	0.6	94(470)	2(80)	£7.88 Add 5% to all orders outside UK.
1	EXP650	3.6	2.4	0.6	46(230)	2(40)	£4.69 All prices and specifications correct
	EXP4B	6.0	1.0	N/A	N/A	4(160)	£3.29 at the time of going to press.

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				2N3441	0.75	2N4121	0.27	2N5295	0.44	40595	0.98	BC147B	0.13		0.15	BC214LB		BC558	0.13	BD245C	0.85	BF185	0.37	BFY52	0.27	MPF3055 1	
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	2N706	0.30	2N2221A 0.25	2N3442									0.13					BCY70	0.21		0.72	BF194					0.44
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	2N708	0.30	2N2222A 0.25	2N3638A		2N4124	0.19	2N5447	0.16	AC126	0.48	BC148C	0.13	BC183	0.12	BC238A	0.13	BCY71	0.26	BD433	0.44	BF196	0.16	BRY39	0.55		0.44
	2N718	0.30	2N2389 0.27	2N3702		2N4125	0.19		0.16	AC1 27	0.48	BC149	0.15		0.12	BC238B	0.13	8CY72	0.18	BD434	0.46	BF197	0.18	BSX19	0.35		0.44
3	2N718A	0.54	2N2369A 0.27	2N3703	0.14		0.19	2N5449	0.20	AC128	0.48	BC149C	0.15	BC183B	0.13	BC238C	0.13	BD115	0.8B	BD435	0.46	BF198	0.19	BSX20	0.35		0.27
4	2N72DA	0.85	2N284B 0.80	2N3704	0.14	2N4284	0.38	2N5457	0.38	AC151	0.43	BC157A	0.15	BC183C	0.13	BC239B	0.16	BD131	0.55	BD436	0.46	BF199	0.19	BSX21	0.35		0.27
1	2N722	0.45	2N2847 1.55	2N3705	0.14	2N4286	0.22	2N5458	0.35	AC152	0.54	BC158A	0.15	BC183L	0.15	BC239C	0.17	BD132	0.75	BD437	0.55	BF224J	0.22	BU104	1.80	MPSA12	
1	2N727	0.50	2N2903 1.60	2N3706	0.14	2N4287	0.22	2N5459	0.32	AC153	0.59	BC1588	0.15	BC183LA	0.16	BC257A	0.18	BD135	0.40	80438	0.55	BF225J	0.27	BU105	1.55		0.33
4.	2N914	0.38	2N2904 0.31	2N3707	0.14	2N4288	0.22	2N5460	0.65	AC153K	0.59	BC159A	0.17	BC183LB	0.15	BC258B	0.19	BD136	0.40	BD529	0.49	BF244A	0.38	BU128	1.08		0.27
1	2N916	0.33	2N2904A 0.31	2N3708	0.12	2N4289	0.22	2N5484	0.37	AC176K	0.70	BC159B	0.17	BC183LC	0.15	BC259B	0.19	BD137	0.41	BD530	0.55	BF244B	0.33	BU204	2.20	MPSA56	0.27
	2N917	0.38	2N2905 0.31	2N3709	0.12	2N4347	2.20	2N5485	0.40	AC178	0.54	BC160	0.38	BC184	0.12	BC300	0.43	BD138	0.41	80535	0.70	BF245A	0.44	BU205	2.40	R2008B	2.45
	2N91B	0.45	2N2905A 0.31	2N3771	2.16	2N4348	2.65	2N5486	0.40	AC187	0.59	BC161	0.38	BC184B	0.13	BC301	0.43	BD139	0.43	BD536	0.70	BF245B	0.44	BU206	2.70	R20108	2.15
	2N929	0.37	2N2906 0.25	2N3772	2.20	2N4918	0.65	2N5490	0.64	AC187K	0.65	BC167	0.13	BC184C	0.13	BC302	0.37	B0140	0.43	80537	0.74	BF257	0.35	BU208	2.70	TIP29A	0.49
	2N929A	0.37	2N29D6A 0.25	2N3773	3.15	2N4919	0.70	2N5492	0.64	AC188	0.54	BC1878	0.13	BC184L	0.15	BC303	0.54	BD181	1.90	BD538	0.77	BF258	0.35	ME0401	0.22	TIP29C	0.65
	2N930	0.37	2N2907 0.25		0.36		0.83	2N5494	0.65	AC188K	0.65	BC1B8A	013		0.15	BC307	016	BD182	2.20	BD539	0.60	BF259	0.35	MF0402	0.22	TIP30A	0.54
	2N930A	0.95	2N2907A 0.25	2N3820	0.39	2N4921	0.54	2N5498	0.67	AD161	1.00	BC188B	0.13		0.15	BC307A	0.16	BD183	2.35	BD540	0.60	BF338	0.42	ME0404	0.17	TIP30C	0.70
	2N1711	0.30	2N2923 0.17	2N3821	0.96	2N4922	0.60	2N6027	0.64	AD182	1.00	BC16BC	0.13	BC212	0.15	BC307B	0.16	B0187	0.95	BDX14	1.32	BF337	0.49		0.22	TIP31A	0.54
	2N1889	0.30	2N2924 0.17	2N3900	0.28	2N4923	0.75	2N6107	0.45	AF106	0.60	BC1698	0.13	BC212A	0.15	BC308	0.16	B0235	0.46	BDX18	1.90	BF338	0.52	ME0414	0.22		0.72
	2N1890	0.30	2N2925 D.19	2N3901	0.30	2N4924	1.15	2N6108	0.55	AF109	0.52	BC169C	0.13	BC212B	0.15	BC308B	0.16	80236	0.44	BDY20	1.10	BFR39	0.30		0.16		0.59
	2N1893	0.30	2N292B 0.17	2N3903	0.20	2N5086	0.30	2N6109	0.55	BC107	0.16	BC177	0.22	BC2121	0.18	BC309A	0.16	BD237	0.44	BDY55	1.90	BFR40	0.29	ME4002	0.16		0.82
1	2N2102	0.50	2N3053 0.25	2N3904	0.18	2N5087	0.30	2N6111	0.49	BC107A	0.16	BC177A	0.22	BC212LA	0.18	BC309B	016	BD238	0.44	BDY56	2.10	BFR41	0.30	ME4003	0.16		0.76
	2N2192		2N3054 0.72	2N3905	0.18	2N5088	0.30	2N6121	0.41	BC1078	0.16	BC177B	0.25	BC212LB	0.18	BC309C	0.16	BD239A	0.44	BF115	0.39	BFR79	0.30	ME4101	0.11		0.97
-			2N3055 0.75	2N3906	0.18	2N5089	0.30	2NB122	0.44	BC108	0.16	BC178	0.22	BC213	0.15	BC327	0.22	BD239C	0.59	BF160	0.33	BFR80	0.30	ME4102	0.11		0.86
				2N4D31	0.55	2N5190	0.65	2NB123	0.48	BC108A	0.16	BC178A	0.25	BC213A	0.15	BC328	0.20	BD240A	0.49	BF161	0.65	BFR81	0.30	ME4103	0.11		1.08
	2N2193		2N3390 0.50 2N3391 0.40		0.65	2N5191	0.75	2N6123	0.45	BC108B	0.16	BC178B	0.25	BC213B	0.15	BC337	0.20	BD240C	0.59	BF167	0.37	BFX 29		ME4104	0.11		0.70
	2N2194			2N4032	0.72	2N5192	0.75	2NB124	0.43	BC108C	0.17	BC179			0.15	BC338	0.23	BD241A	0.49	BF173	0.37	BFX30	0.34	ME6101	0.22		0.59
	2N2194/		2N3391A 0.45	2N4036						BC109C		BC179	0.25	BC213C	0.17	BC547	0.13	BD241C	0.65	BF177				MEG101	0.22		1.05
	2N2195		2N3392 0.17	2N4037	0.60	2N5193	0.75	40361	0.55		0.16		0.25	BC213L			0.13	BD242A	0.55		0.27	BFX84	0.30				0.50
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1	2N2217		2N3394 0.17	2N4059	0.17	2N5195	0.97	40363	1.45	BC109C	0.18	BC179C	0.26			BC5478	0.13	BD243A	0.62	BF179	0.33	BFX86	0.30	MJE340	0.62		
	2N2218		2N3395 0.19	2N4060	0.22	2N5245	0.37	40408	0.82	BC140	0.30	BC182	0.12	BC213LC		8C548	0.13		0.65	BF180	0.37	BFX87	0.35	MJE370	0.62		0.22
	2N2218/	4 U.38	2N3396 0.19	2N4061	0.19	2N5246	0.38	40409	0.82	BC141	0.32	BC182A	0.12	BC214	0.17	BC549	0.14	BD243C	0.87	BF181	0.37	BFX88	0.30	MJE371	0.86	TIS91	0.27

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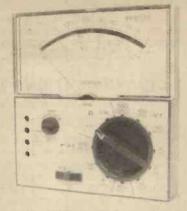
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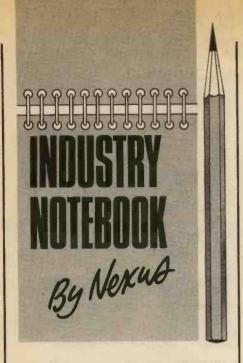
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MATS

There's nothing quite like having a telephone in your car linked into the public network. Once a toy for the affluent it is now appealing to ordinary business people.

In the UK the service is manual. That is, you need to call up by radio from your car and an operator will connect you to the telephone number you want.

A big advance on this system is now appearing in the Middle East whereby you can dial the number directly from your car to any subscriber in the world who can be accessed through local or international STD. Similarly a fixed or mobile subscriber can call a remote mobile just by dialling his number and be connected automatically without operator intervention.

Cable & Wireless engineers have recently engineered such a system for Bahrain Telephones and the Mobile Automatic Telephone System (MATS) went into service on April 25 with a capacity of 50 mobile subscribers with facilities to expand the service to as many as 1,500 subscribers. A similar system has been operating in Qatar but the Bahrain system is said to use more advanced technology.

There seems to be an insatiable demand for communications in the whole of the Middle East. Bahrain, of course, is now emerging as a major business centre for the area, displacing Beirut where political problems and civil disturbances have forced many companies to look elsewhere for a secure operating base.

So great is the demand for communications and the ensuing business for equipment suppliers that Bahrain is to have its own communications exhibition and conference. It is called MECOM 79 and opens on April 23 next year. Organised by a British exhibition

company, it is sponsored by the Bahrain Ministry of Communications and Transport and also has the support of the PTT administrations of Saudi Arabia and Oatar

The associated technical conference has the support of the Bahrain Society of Engineers and the British journal Communications International.

COMMUNICATIONS 78

From looking forward we now look back on Communications 78 which, this year, moved from Brighton to Birmingham. I didn't welcome the move and I was not alone in preferring the civilised pleasures of Regency Brighton to those of Britain's second largest city. But arriving for the show on the first day it was quite obvious that the facilities at Brighton's Metropole exhibition centre could not have physically accommodated the enormous expansion since 1976.

So we suffered in silence in the courtesy coaches from the remote car parks to the exhibition hall and the tenminute trudge in chilling gales from the hall to the conference centre, knowing it was in a good cause and finally dispelling any doubts that communications is the great growth area of this decade and, for that matter, decades ahead.

The event was a triumph for Tony Davies, the organiser, and his committee drawn from Government departments, the armed forces and from industry, right through from the inaugural luncheon graced by the Duke of Kent to the final attendance count of over 15,000 registered visitors from over 40 countries, more than twice as many as attended the 1976 event. The conference, too, broke all records with 825 registered delegates and proceedings which ran to more than 400 pages of condensed print.

Although not a direct result of the exhibition, some exciting contracts were announced at the same time. One was a multi-million pound contract from Saudi Arabia awarded to Cable and Wireless to equip the para-military Saudi National Guard with a communications system. The other was awarded to Marconi Space and Defence Systems who are to develop, in conjunction with Cincinatti Electronics, the next generation of v.h.f. mobile radio equipment for the United States Army, the Sincgars project.

Racal and Plessey were also in the bidding for the US Army contract, both with American partners. It was understood at one time that of all those bidding only one winner would be selected. In the event the Americans selected two winners (the other being ITT) and each will have to produce evaluation models for a final play-off to determine the ultimate winner who then gets the jackpot of orders which some observers estimate as being worth 500 million pounds.

Meanwhile the Ptarmigan tactical trunk communications system for the British Army is forging ahead with Plessey as main contractors. A Plessey spokesman told me that the project was on target both in performance and timescale and, in fact, system hardware was on public show for the first time and is already in the hands of the users who describe the system as a "healthy bird". Project development started in 1973 on a seven-year time scale. Plessey also introduced two new military radios at the show, both private venture, claimed to fill gaps in the present Clansman range for military use.

HANDS OFF!

When Racal Electronics Group purchased Milgo of Miami, Florida, last year they also bought some outstanding litigation which involved patent infringement against United Business Communications Inc. The company, wholly British-owned and re-named Racal-Milgo Inc., has received awards of about three million dollars by the US District Court in Kansas in court costs, damages and interest since 1972. Three patents were involved and the infringement, according to Senior Judge George Templar, was so serious that he ruled that Racal-Milgo were entitled to treble damages.

INVESTMENT OVERSEAS

British companies, still nervous of industrial unrest, creeping nationalisation and inflation at home, are continuing to search for better returns overseas. Thorn, for example, has just bought two US companies, spending six million pounds in the process. Both will complement the Thorn Industrial Control Engineering Group. This at the time when Thorn was facing the prospect of strikes at the company's Yorkshire TV factories over proposed redundancies.

And, Racal, always on the look-out for acquisitions has gathered in California-based Vadic Corporation In the low-speed modem business and its first French company, Dana Electronics France SA. Vadic complements the high-speed modem business of Racal-Milgo giving Racal a complete modem product range, while Dana in France is the tail-end of the purchase of Dana Laboratories Inc. of California, the acquisition of which was announced last October.

Racal paid nearly 5-4 million pounds for Vadic Corporation (now to be known as Racal-Vadic Inc.) which has only 200 employees producing a profitable turnover of five million pounds per annum and expanding fast. This is a turnover of 25,000 pounds per employee representing a level of productivity which good business demands and expects, and it well illustrates the sinking labour content of modern electronic equipment.



An accurate wide range timer with audio and visual alarms

The electronic timer described here was developed to meet two requirements: the need for a general purpose twelve hour timer capable of switching on an external load (low voltage d.c. to 240V mains a.c.) at the end of preset time periods; and also capable of enabling a cine camera to take time lapse sequences so that observations could be made of growing plants, etc.

For the latter technique the timer is able to switch on a flood lamp at the end of a preset period and then to release a camera shutter while the subject is illuminated; the timer then resets itself automatically for another sequence.

The meeting of these two needs has resulted in a very versatile timer which has the following specification:

- (a) Mains or battery operation.
- (b) Battery operation provides a portable timer (with optional audio alarm) which can be used to switch on low voltage battery-operated equipment.
- (c) Mains operation provides, in addition to (b), triac control of mains powered loads up to 1,000W.
- (d) In both battery and mains modes of operation, "one shot" or "repeat" timing can be selected.
- (e) Two linked 12-way wafer switches select 11 delay periods of 1 hour and 11 delay periods of 5 minutes giving a maximum delay of 11 hours 55 minutes in 121 intervals of 5 minutes.
- (f) A range switch divides the above intervals by factors of 10, 100 and 500 giving, on the "1 hour" wafer switch, intervals of 6 minutes, 36 seconds and 7.2 seconds respectively; and on the "5 minutes" wafer switch intervals of 30 seconds, 3 seconds and 0.6 seconds, respectively.
- (g) When the timer is operated from the mains and is in the "repeat" mode, the triac drive circuitry can be used to switch on a flood lamp (max 1,000W) for 2.5 seconds. During this period, a low voltage pulse is available of about 0.5 seconds duration for driving

- a solenoid to operate a camera shutter. The lamp then goes out with the timer having already set itself for another cycle. In the "one shot" mode, once mains switching is initiated the load remains energised.
- (h) Repeatability of any time intervals is to within 0.01 per cent but the accuracy of the preselected intervals depends on the care with which the constructor selects the values of 22 resistors and 3 capacitors and upon the stability of these components.

CIRCUIT OPERATION

The complete circuit diagram of the unit is shown in Fig. 1. IC1 is a precision digital timer which makes use of an "on chip" oscillator the frequency of which is set by an externally connected CR network. Pulses from this oscillator are fed through a twelve stage binary counter which switches the output stage after 1,095 counts.

Timing is started by closing S4 to provide power to the timer, the time period being set by the values of R22 to R32 and R33 to R43 which are in series and selected by the wafer switches S5 and S6, and the values of the capacitors C6, C7 or C8 selected by S7. This CR combination across pins 13 and 14 sets the frequency of the internal oscillator which times out after a time T given by T=2736 CR.

During the count down period, pin 3 of IC1 is high, allowing D2 to light. At the end of the count down, output pin 2 goes "high" and pin 3 goes "low": D2 goes out and TR1 is switched on. What happens subsequently depends upon the settings of the various switches.

Firstly, if S1a is open, the voltage at pin 3 stays "low" and TR1 remains switched on: if S1a is closed, R2 and C5 ensure that the timer automatically resets after about one second via the trigger of IC1 (pin 1). The period "t" before resetting takes place is given by t = 0.6 R2 C5, approximately.

If S3 is closed, the audio alarm signal of IC2 is switched on

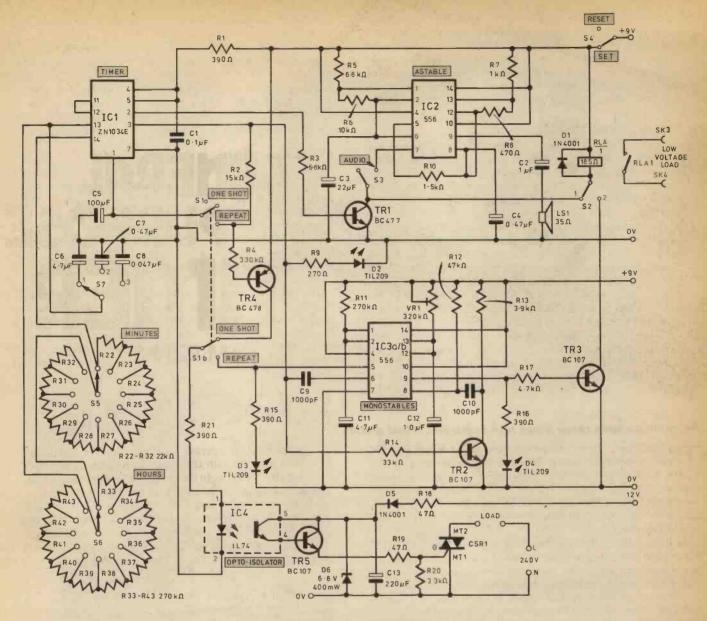


Fig. 1. Circuit diagram of the General Purpose Timer

whilst TR1 is on. This dual "555" timer (IC2) is operated as a dual astable multivibrator which is d.c.-connected via R10 so as to produce a repeated "pip" from the speaker. With or without S3 closed, transistor TR1 can energise relay RLA if S2 is in position 1. In the "one shot" mode, RLA remains energised and can be used to switch on a low voltage load (not mains) for an indefinite period at the end of the timing period. In the "repeat" mode, TR1 is energised for about one second so the alarm and RLA are energised briefly at the end of the time period.

If S2 is in position 2, RLA is energised via a positive pulse appearing at pin 9 of IC3. Once again a dual "555" timer integrated circuit is used for IC3 but each individual timer is wired as a monostable multivibrator. These two monostables are arranged to be triggered one after the other by the square wave pulse appearing at pin 3 of IC1 when operating in the "repeat" mode. But the trigger pins 6 and 8 of IC3 require negative-going trigger pulses via capacitors C9 and C10, respectively, in order to initiate delays determined by the values of R11/C11 and VR1/C12 respectively.

When the voltage at pin 3 of IC1 rapidly goes negative at the end of the timing period, monostable IC3a is triggered for a period of about 2.5 seconds; this negative going pulse keeps transistor TR2 turned off allowing pin 8 to remain "high" via R12. However, if IC1 is in the "repeat" mode, one second after IC3a has been triggered pin 3 goes "high" once again for another timing period "T".

This rapidly rising positive voltage switches on TR2 so that the sharp rise in current through R13 provides the required negative pulse at the collector of TR2 which, applied to pin 8 by C10, triggers the second monostable IC3b. Thus a positive pulse of about ½ second duration appears at pin 9 approximately one second after the positive pulse has appeared at pin 5.

The "on" states of the monostables are indicated by diodes D3 and D4. Transistor TR3 enables RLA to be energised for about 0.5 second if required so that its normally open contacts can be used for providing solenoid drive for a camera release button.

Note that RLA can be energised by either the positive

pulse from IC1 pin 2 (for one second or for an indefinite period at the end of the timing period depending upon the position of S1), or by the positive pulse at pin 9 of 1C3b thereby not being energised when IC1 is operating in the "one shot" mode or for one second after IC1 has timed out if in the "repeat" mode. This latter pulse lasts only long enough to provide camera shutter drive through a solenoid.

TRIAC OPERATION

Whether the triac provides switching for a mains load over an indefinite period or for about 2.5 seconds at the end of the timing period depends on the position of S1b. In the "repeat" mode, S1b directs the 2.5 second pulse from pin 5 of IC3a to the opto-isolator IC4 to energise the load.

At the end of this period the triac automatically switches off this a.c. load. In the "one shot" mode, S1b directs the negative going pulse from IC1 (pin 3) to pin 1 of IC4 (after inversion by TR4) in order to switch on the load for an indefinite period. Transistor TR4 was found necessary to ensure that pin 3 was not too heavily loaded for this caused erratic operation of IC1 in the "repeat" mode. Note that TR4 is base biased by R4 just sufficient for its collector current to bring on the internal l.e.d. of IC4.

The one shot operation of the timer means that mains operated appliances can be switched on after any one of the preselected time periods up to a maximum of 11 hours 55 minutes. The mains load can be switched off using either the "set/reset" switch or, of course, by switching off the mains.

AUDIO ALARM

This can be switched in and out by means of S3 and is based on IC2; it can be operated in both the mains and battery modes. In the "repeat" mode operation of timer IC1, a one second burst of "pips" is heard at the end of each delay period. In the "one shot" mode the pips sound continuously until the timer is switched off.

The alarm will work on battery or mains; care should be taken to ensure that, when the timer is operated using a battery, it does not become discharged during long preset delay periods.

POWER SUPPLY

The circuit diagram of the p.s.u. is shown in Fig. 2. Switch S8 selects mains or battery operation. Approximately 9V is derived from the mains using one winding of the transformer

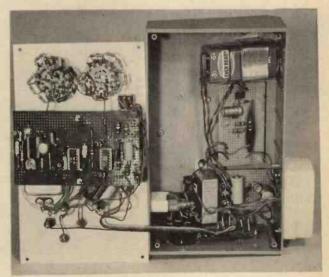
and the bridge rectifier (D8 to D11). This voltage is stabilised by the Darlington series regulator.

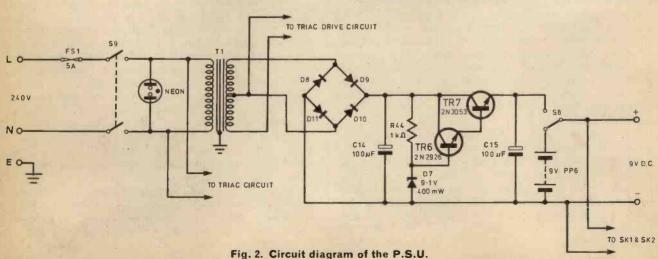
The other winding provides the supply for TR5 on the triac drive board and for the internal phototransistor of IC4. The 12V a.c. from the transformer is rectified by D5 before being applied to the collector of TR5. The Zener diode D6 and the electrolytic capacitor C13 ensure that the voltage is sufficiently steady for reliable operation of the triac. It is the rising emitter voltage of TR5 which switches on the triac.

CONSTRUCTION

Three pieces of 0·1 inch matrix stripboard were used for the circuit assembly: the main timer board, the triac drive board and the low voltage board. The component layouts for these boards are shown in Figs. 3, 4 and 5. All external connecting wires to the boards should be made longer than required and then cut to the correct length prior to being soldered to the l.e.d.s and switches. Note that the l.e.d.s have polarised leads and care should be taken to ensure correct connection to the circuit board.

The timer was housed in a polystyrene case with a sloping front panel. The aluminium panel was removed and replaced by a polystyrene one to avoid the necessity of having to earth it although no mains connection is made to this panel. The stripboard circuit of the timer is attached to this control panel by stand-off studding to clear the switches and the l.e.d.s, etc.





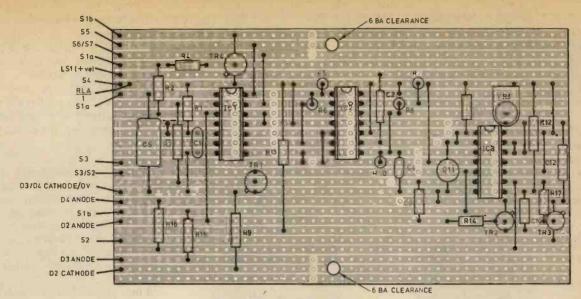


Fig. 3. Main timer board layout

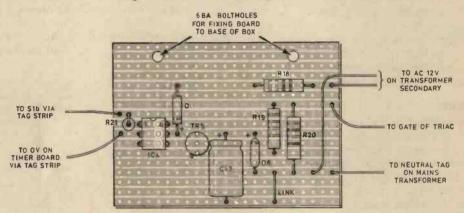


Fig. 4. Triac drive board layout

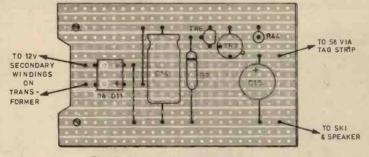


Fig. 5. Low voltage board layout

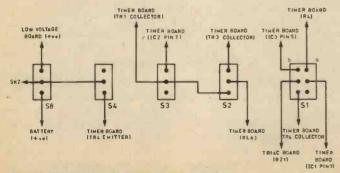
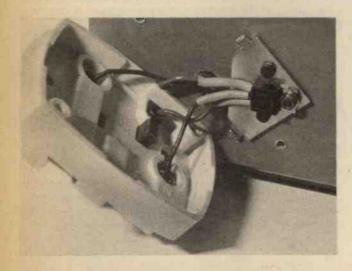


Fig. 6. Wiring of the front panel switches



R.S. type
R.S. type
R.S. type
R.S. type
R.S. type
pe 261-801
gle
gle (4 off)
W. L. S. L. S. C. S. C. S.
30 ≥ 79mm
ALC: NAME OF STREET
THE RESERVE
d socket
0.5A
e holder
Burn State Co.



A solder tag attached to one of the terminals of the speaker is used as a common low voltage negative connection. The wiring of the switches S1, S2, S3, S4 and S8 is shown in Fig. 6.

The triac drive board is bolted to the bottom of the box adjacent to the low voltage board. The component layouts in Figs. 5 and 6 show the external connections which are to be

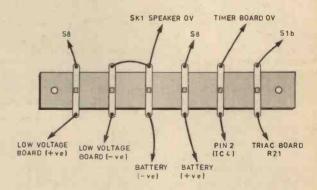


Fig. 7. Barrier strip wiring

made to these two boards. These connections should not be made to the timer board until the latter has been tested.

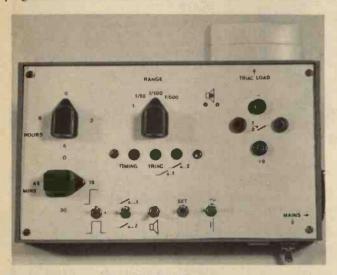
The mains switch (S9) is mounted on the front of the box and isolates both live and neutral wires when off. Also, with the switching arrangements employed, the battery may be switched in by S8 to control timing functions and to provide triac drive whilst mains loads are switched via the triac.

The triac was mounted inside the base of the surface

mounted 13A load socket on a small piece of aluminium which acted as a heat sink.

In order to facilitate the fitting and removal of the control panel, a six-way length of barrier strip was bolted to the side of the timer box adjacent to the transformer. In this way the wires from the control panel can be routed neatly to the power supply circuits and the triac control panel. The leads requiring connections to this barrier strip are shown in Fig. 7.

The mains connection to the unit should be made via a 3 pin non-reversible plug and socket. In the prototype the plug was fitted in the side of the case near the mains switch.



SETTING UP AND TESTING

When the timer board has been assembled and mounted on the control panel and before connections are made to the low voltage and triac drive boards, the timer circuit can be connected to the 9V battery and tested. Switch S8 should be in the "battery" position and the audio alarm switch S3 "on". Set switch S5 (the "hours" switch) to zero and S6 (the "minutes" switch) to 5 minutes—the first position clockwise.

The range divider switch S7 should be set to divide this 5 minute period by a factor of 100. Now operate the "set/reset" switch to provide power to the timer circuit and the green l.e.d. (D2) should light and go out after 5 minutes × 60/100 seconds or 3 seconds. Immediately the l.e.d. goes out indicating the end of the timing period, the audio alarm should sound and the red l.e.d. (D3) should light up for about 2.5 seconds before going out.

What happens after this sequence depends on the setting of S1. With S1 in the "repeat" mode, D2 will relight about I second after it has gone out at the end of the timing period, i.e. the timer has reset itself for another delay period; and the audio alarm stops. As soon as D2 relights, D4 lights briefly indicating that a solenoid can be operated to drive a camera button if S2 is in position 2.

This sequence of switching will repeat every 3 seconds but can be altered to any other delay period as described at the beginning of this article. With S1a in the "one shot" mode, D2 will go out and stay out, D3 will come on for 2.5 seconds and go out but D4 will not light.

Simultaneously with the above sequences, the low voltage relay will be heard to operate. With S2 in position 1 and S1a in the "repeat" position, the relay will energise as D2 goes out at the end of the delay period (closing the contacts of the relay) and de-energise as D2 comes on again: the relay energises for about one second at the end of every delay period.

With S2 in position 2 and S1a in the "repeat" mode, the relay energises for the time S3 is alight thus providing automatic switching for a camera shutter drive. With S1a in the "one shot" mode, and S2 in position 1 the relay will energise at the end of the time and remain energised.

If S2 is in position 2 the relay will not energise and TR2 will not be switched on since the monostable based on IC3a is not triggered. These sequences are summarised in Table 1.

The low voltage board should be checked to see the voltage available from the circuit is about 9.0V. If it is, the leads from this circuit board can be connected to the timer board

SWITCH S1a	SWITCH S2	RELAY RLA
"Repeat"	Position 1	"On" at end of delay period (D2 goes out)
		"Off" one second later as timer resets (D2 comes on)
"Repeat"	Position 2	"On" as D4 lights
		"Off" as D4 goes out
"One shot"	Position 1	"On" at end of delay period and stays on
"One shot"	Position 2	Does not energise

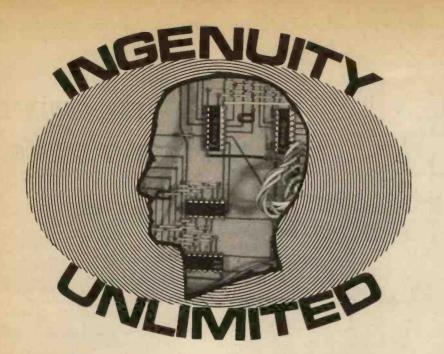
Table 1. Relay operation for the different switching sequences of S1a and S2

via the barrier strip, the negative lead to the solder tag on the speaker and positive lead via switch S8 to the timer board. The whole sequence of timing operations described in Table 1 can also now be checked for correct operation from the mains.

Finally the triac drive board should be tested. The lead from S1b should be connected to R21 on the triac drive board and the lead from pin 7 of IC1 on the main timer board to the 0V line and pin 2 of IC4 in the triac drive board also to the 0V line. With the timer being operated from the battery, check that the opto-isolator is working as follows. Set the timer to operate in the "repeat" mode every 9 seconds, say. Use a multimeter set to the "Ohms \times 1k" range and connect the black lead to pin 5 of the IC4 and the red lead to pin 4 of the same i.c.

Now every time monostable IC3a is triggered and D3 lights the resistance between these two terminals will fall to a low value. This connection is actually across the collector-emitter connections of the internal phototransistor of the opto-isolator, and when the resistance between these connections falls it indicates that it is receiving light from the internal diode.

The energising of this phototransistor switches on TR5 and the triac when the triac drive board is connected to the mains. Now switch S1b to the "one shot" position and note that the resistance across pins 4 and 5 of IC4 falls at the end of the delay period and remains low. If all is well the triac drive board can be wired to the neutral and the 3 pin load socket.



A selection of readers' original circuit ideas. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.

Why not submit your idea? Any idea published

will be awarded payment according to its merits.

Articles submitted for publication should conform to the usual prac-tices of this journal, e.g. with regard to abbreviations and circuit symbols. Diagrams should be on separate sheets, not inserted in the text.

Each idea submitted must be accompanied by a declaration to the effect that it is the original work of the undersigned, and that it has not been accepted for publication elsewhere.

WHEN working with digital systems, the designer often has need of a cheap TTL compatible PROM. One very interesting device released is the 256×4 bit DM8574 by National Semiconductors Ltd. This i.c., with the exception of certain drive requirements, is pin for pin compatible with the popular 16 pin d.i.l. 2112 256 × 4 bit RAM, whose R/W pin is replaced by a CE input on the PROM. It is therefore quite simple to design a general pur-pose memory board for most micro-processor systems which mixes RAM and ROM in any combination on the same printed circuit board.

The only problem with the use of PROM is the need for a device programmer. Fortunately, the DM8574 is a "Field Programmable" device of the fusible link variety and a simple programmer can be constructed as shown.

The PROM may be programmed by supplying a single 5V supply to the device and selecting the address of the 4 bit word which is to have some of its fusible links blown. (Blowing a link creates a logical zero at the bit address selected; leaving the link unblown leaves a logical one at that position). A logical one is then applied to both the enable inputs, and the programming pulse (see figure) is fed to the output where a low level is required. Verification that the bit has been programmed may be obtained by bringing the enable inputs low and checking the logical level at the appropriate output.

appropriate output.

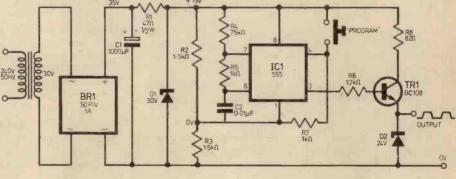
The programmer utilises a 555 timer working from a 15V supply derived from a split 30V supply so that when the 555 output goes high the base of the BC108, whose emitter potential is held at 24V, is driven

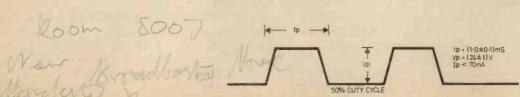
hard on, causing some 70mA to flow from its output. The 555 is arranged to provide a 1kHz 50% duty cycle square wave at its output when pin 4 on this i.c. is brought high by pressing the "PROGRAM" push-button, this output then being used as explained to drive the BC108 when programming the PROM bit-by-bit as explained.

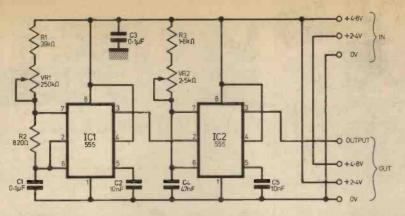
Microcomputer addicts might like to consider the possibility of automating the programming sequence in order to provide a quick and easy way of obtaining custom-programmed ROM chips suitable for inclusion in their system.

J. A. Murdie, Darlington, Durham.









RADIO CONTROL SERVO TESTER

THE servo mechanisms currently used in radio controlled models use a variable width pulse to convey the information concerning the required position. This pulse has been standardised as varying between 1-2ms and is nearly always positive going. Most receivers and decoders use 4-8V or 6V batteries and the amplitude of the pulse is almost equal to the supply voltage.

The information for each channel of a multi-channel system is transmitted sequentially, followed by a synchronising pulse to maintain lock between receiver and transmitter. Models using several channels are common-

place; aircraft often use six.

When setting up the control systems in a model, it is usual to have to use the transmitter and receiver together. This can be inconvenient especially at a flying field where spurious signals are most definitely not wanted! This

servo tester overcomes the need for using all the gear just to test one function by simulating the decoder output for one channel only. It is simply connected between the battery and the required servo. The pulse width is adjustable over the full range, and the repetition rate may be set for any number of channels, although this is not really necessary and this adjustment may be omitted. (Servos work in equipment of one to seven channels without adjustment, and the variable rate function was built in when servos were being designed, purely to check this fact.)

The circuit consists of an NE555 wired as an astable pulse generator, producing narrow, negative going spikes at a repetition rate of between 3 and 20 ms. A good compromise, to eliminate one adjustment, would be 10 ms. This is achieved by replacing R1 and VR1 with $100k\Omega$. The narrow

pulses produced trigger a second 555 wired as a monostable, the duration of which is adjustable between 1 and 2ms. The internal design of the 555 is such that varying battery voltages do not affect pulse width so that the circuit will work from 4 to 15V supplies.

The output pulse is almost as great

as the supply voltage.

Both circuits are assembled on commercially available circuit boards. It is not necessary to calibrate the device, the central position of the width control potentiometer will be very close to 1.5ms, the mid-range value. It may be checked against a new servo; here the end positions should be at the limits of useable travel, usually ±45°. The centre tap from the battery is not necessary for the working of the device, and is included as some of the older servos require it.

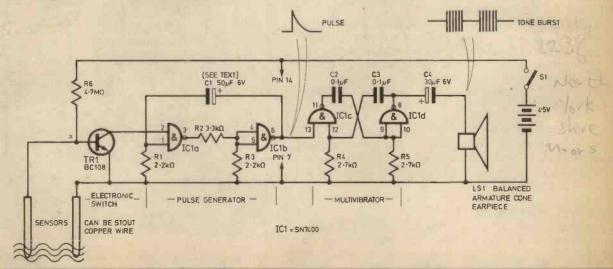
A. Langton, Aberdeen.

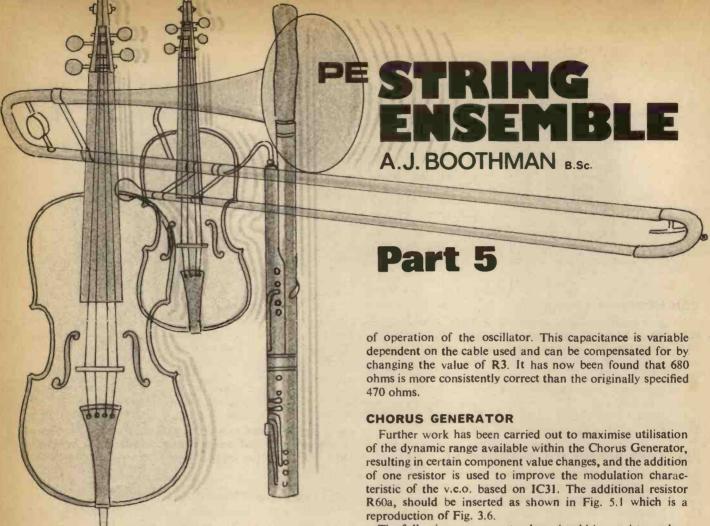
OVERFLOW INDICATOR

GATES IC1a and IC1b form a pulse generator. C1 determines the frequency of the pulses, and the value of this should lie between 10μ F and 50μ F, the larger the value the longer

the duration of tone. Gates ICIc and d together form a multivibrator to produce the required tone. The multivibrator is controlled by the pulse generator through pin 13, so when the generator is not producing pulses, no tone is heard from the transducer. TRI forms the switch that is required to control the pulse generator. When water or any conductive liquid bridges the sensor probes, bleeps are heard from the earpiece.

V. J. Guillaumier, Malta.





HE final part of the series gives details of the cabinet construction, a number of component value changes for improved optimum results, and further interwiring information and clarification.

CABINET CONSTRUCTION

The component parts for the cabinet are given in the cutting list with assembly details shown in Fig. 5.2. Some variation in dimensions may be required to suit alternative keyboards, but the construction has been kept simple for ease of adaptability. Corner joints can be made from short lengths of §in (15.875 × 15.875mm) timber, screwing from the inside of the cabinet.

IMPROVEMENTS ON PROTOTYPE

Since commencement of publication of this series the construction of further models has resulted in information from which certain component value changes are recommended to give the best results.

TONE GENERATOR/PSU

Connections from the p.c.b. to the front panel transposer switch and the tuning potentiometer are made with screened cable (see Fig 5.2), and the capacitance of the cable contributes to the capacitance C8 which determines the frequency

The following component values should be used to replace those given in Part Three of the series. R30/44-10kF $R16-27k\Omega$ $R22/36-6.8k\Omega$ $R65/66-220\Omega$ C16-470pF $R59-15k\Omega$

R60a-39kΩ C46-680pF

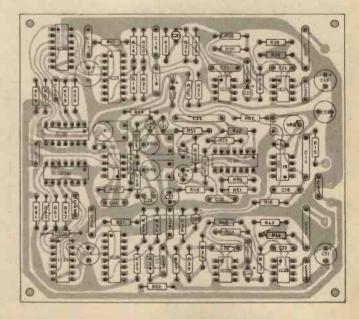
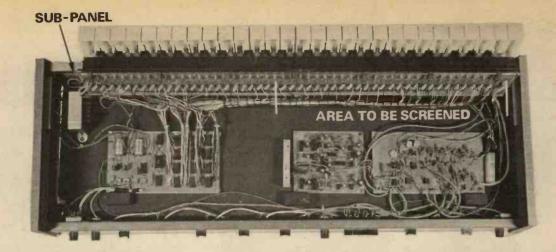


Fig. 5.1. Showing position of additional resistor on chorus p.c.b. Orientation of the I.c.s is also given



VOICE/PREAMP CIRCUITRY

Noise contribution from the Woodwind and Brass filters has been reduced to negligible proportions by replacing R107 and R116 with shorting links, removing C76 and C83 and connecting $1M\Omega$ resistors across pins 2 and 6 of 1Cs 33 and 34.

The following components on this figure should also be changed as follows:

R124—150k Ω , R131—470k Ω , R135—68k Ω C81—47nF, C88—47nF, C101—220pF

PSU/TONE GENERATOR INTERWIRING

Last month an interwiring diagram (Fig. 4.6) was given for the Chorus Generator and Voice/Preamplifier p.c.b., which included details of most of the front panel control connections. Fig. 5.2 shows the wiring for the remaining controls, which are connected to the P.S.U./Tone Generator p.c.b., and the mains input sub-panel.

The sub-panel is mounted at the rear of the cabinet, at the top left-hand end (see photo) and contains the mains switch, fuse and socket which are wired to the mains transformer as shown. Earth leads are also required both to the keyboard chassis and 0V pin on the Voice/Preamplifier p.c.b.

Low voltage a.c. connections are made in pairs between the transformer and the p.c.b. supplies +15V, 0V, and -15V are taken direct to the Voice Board, and can be routed behind the key bar to avoid entanglement with the Tone Generator output leads.

Connections to D13 on the front panel are shown, the polarity being indicated by the "flat" on the skirt of the l.e.d. The lead adjacent to the flat should be wired to the l.e.d. pin on the p.c.b. D13 is fitted adjacent to the pitch control. The 20V supply is wired to Attack and Sustain Controls as shown in the diagram and connections are made to the keyswitch rod and sustain line on the Diode Gate assembly from these controls.

TRANSPOSER/TUNING CONNECTIONS

A four core screened cable is used to connect the Tone Generator p.c.b. to the Transposer Switch, the screen is connected to the ground terminal on the p.c.b., but not to the switch. A single wire link should be connected between the switch and tuning potentiometer VR5, and a screened cable between the potentiometer and the p.c.b.

In this case the screen should be soldered to the p.c.b. at one end and the body of the potentiometer at the other end.

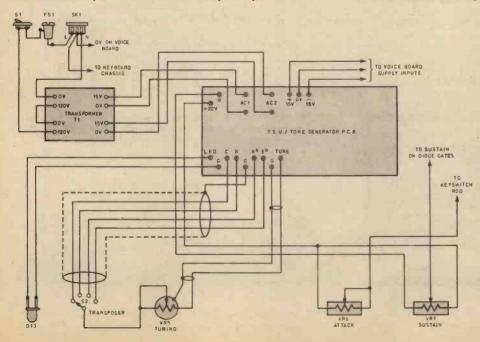


Fig. 5.2. Interwiring of P.S.U./Tone Generator to controls

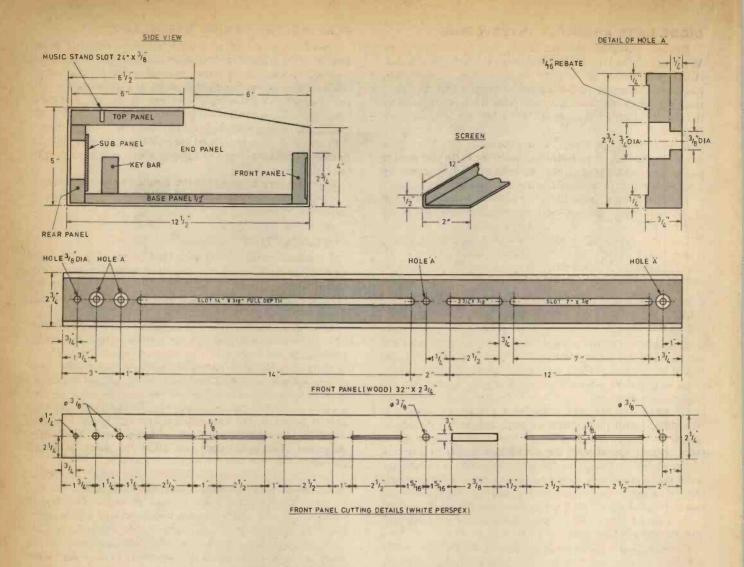
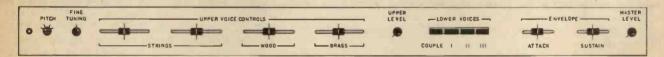
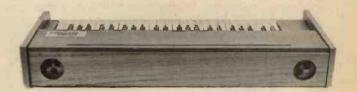


Fig. 5.3. Cutting details of casing and fascia



CUT	TING LIST			
End Panels (2 off)	12½in × 5in × ½in Veneered			
Base Panel	32in × 10½in × ½in Chipboard			
Rear Panel	32in × 3\frac{3}{2}in × \frac{3}{2}in Veneered			
Front Panel	32in × 2⅓in × ⅓in Veneered			
Top Panel	32in × 6in × ≩in Veneered			
Key Bar	$22in \times 1\frac{7}{8}in \times \frac{7}{8}in Timber$			
Sub-Panels (2 off)	4in × 3in × 1/8 in Plywood			
Keyboard Ends (2 off)	To suit Keyboard			
Metal Screen	12in × 2½in Bent as shown in			
(Fig. 5.3)				
Sub-Panels are mounted against 2in holes in Rear Panel				

Fig. 5.4. (Above) Showing fascia legends. (Below) Showing rear panel and attached sub-panels. Right is for mains input and left for output sockets



DIODE GATE ASSEMBLY INTERWIRING

The interwiring between the Diode Gate assembly and the Voice Board, Sustain control and earth is shown in Fig. 5.2. A lead is connected between the point at which the earth lead has been connected to the keyboard chassis and the earth track on the Diode Gate assembly. This track is identified as the one to which all capacitors C13 (4.7μ F electrolytics), are soldered.

In assembling the Diode Gate boards earlier, links were made between the three boards to make the earth and sustain tracks continuous (Fig. 2.6). The sustain track is joined by a single wire to the Sustain Control VR7.

The four output track links are made between PCB2 and PCB3 only since the keyboard split occurs between PCB1 and PCB2.

Low and high output groups are each connected to the Voice Board using a four-core screen cable, with the screen connected at the Voice Board end only. The low output cable should be routed behind the key bar.

GATE ASSEMBLY—TONE GENERATOR WIRING

The interconnection pattern for the Diode Gate p.c. b.s was given in Fig. 2.8 in the second part of the series. This illustrated the three p.c.b.s, making up a single assembly, with arrows indicating the 85 points at which leads are taken from the Diode Gate assembly to the Tone Generator board. The track leads associated with the connection points can be seen in the single board photograph on page 596, and if the photograph is turned through 180° the four rows of leads can be seen to correspond with the connection points in Fig. 2.8. As described in Part Two the interconnection is carried out with easily solderable insulated copper wire; a single length acting as both the interconnecting wire on the Diode Gate assembly and the flying lead to the Tone Generator board.

In Fig. 2.8 the numbers in circles (1-8) correspond with the numbers on the outputs of IC7-18 (1-8) on the Tone Generator p.c.b. shown in Fig. 1.7 in the First Part of the series

Each arrow represents a lead back to the Tone Generator with the note identified by checking the vertical correlation with the letters below the arrows and the circled number under the row concerned. On the left-hand end the circled number changes for the last note in a row, as also occurs in the case of C5.

SCREENING

In the design of the String Ensemble great attention has been paid to ensuring that the instrument is free from beehive (background noise from the Tone Generation circuitry). This has been achieved in respect of blocking any direct beehive through the Diode Gates into the voice circuitry. However, it is important to ensure that this is not negated by pick up into the voice circuits from the Tone Generator harness. The use of screened leads where recommended will prevent this, and care should be taken not to strip the screening back too far, leaving unscreened lengths, particularly in connections to the Woodwind slider potentiometer.

An earthed metal screen is also required to isolate the Tone Generator harness from the Voice and Chorus boards when the keyboard is lowered, and whilst the original prototype incorporated a screen fixed above the boards on stand-offs from the baseboard (see photo on page 598), it has now been found more convenient to mount the screen onto the keyboard chassis with equal success. Dimensions for the screen are given in Fig 5.3.

PLAYING THE STRING ENSEMBLE

Prototypes have now been used on a wide range of power amplification systems in both domestic and stage settings. A normal hi-fi amplifier/speaker combination has given excellent results with a power handling capacity of approximately 15W per channel using the high level output for the Ensemble into the auxiliary channel of the amplifier. With a commonly used guitar combo-amplifier rated at 100W the "Bright Input" with treble lift, has been used successfully to give either solo or chordal backing effects in a group which includes an organ for general use.

With this type of amplifier the Ensemble low level output is recommended to prevent overloading of the first stage of the combo-amplifier.

LEVEL SETTING

The Upper Voice controls plus the Upper and Master Level controls give six variable signal levels which when coupled with the Lower Voice switches and the variation in the number of notes playing at any time results in a very wise signal range driving into the chorus delay lines.

For a good signal/noise performance it is important to set the instrument such that the delay lines have a reasonable signal level, but due to the wide range of signal available it is possible to overload the delay lines under extreme conditions.

The new component values given earlier for the Chorus and Voice boards give a good optimum condition where in normal playing circumstances the Upper and Master Level Controls should be set in the mid-position and the power amplifier adjusted for maximum output with the swell pedal fully depressed. The pedal does not affect the delay lines and should be used for normal volume adjustment.

A single Upper Voice can be used with the slider control at maximum. Two voices can be used with each slider control in the mid-position. Similar compensation can be made if more voices are mixed, but since an overload safety factor has been incorporated it will probably not be necessary.

An alternative mode of operation when mixed voices are to be used regularly, organ simulation, is to turn the Upper Level control anticlockwise to, say, the one third position which will then allow all voices to be mixed additively with heavy chordal fingering without overloading the delay lines.

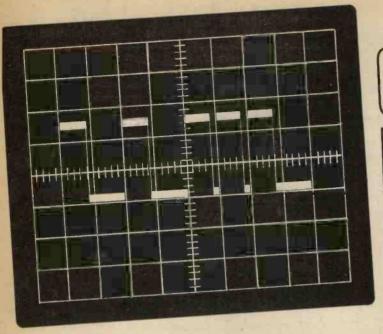
For solo playing on single voice the Upper Level control can be turned to maximum. The Master Level controls the entire keyboard and can be used to increase the lower keyboard level, whilst the upper keyboard level is reduced by the Upper Level control, or to further increase the upper keyboard level on single note solo playing.

OPTIMUM VOICING

The best effects are obtained with simple voice settings, that is with one voice control in use at any particular time. However, experimentation with the wide range of controls available can lead to the creation of other sounds which with subjective opinion can enhance the motives of an individual to construct the project.

Constructors who would like a cleaner copy of Fig. 3.5 should write in including a stamped addressed envelope. In Fig. 4.3 and the Components List last month C81 and C82 should be 4,700pF and 47nF respectively.





GR.O. LOGIG MONITOR GENERATOR

R. DERRY

THE device to be described uses two MSI TTL logic chips with a minimum of additional components to generate an analogue waveform which gives a histogram type of display when fed to an oscilloscope. In addition, marker pulses can be added to the display to add to its clarity. An eight channel version is described which can be built for about £5, assuming that an oscilloscope is already available.

Also described is a way of using the display to test logic displays which is particularly useful for general bench work with logic circuits. The basic design can be extended to cope with more than eight channels if required.

CIRCUIT DESCRIPTION

The heart of the circuit is a 74151 i.c. as seen in Fig. 1. This is an eight line to one line data multiplexer operated conventionally being addressed by a divide by eight counter (a 7490 counter wired to divide by eight). The counter is driven from an asymmetrical oscillator fed via a Schmitt trigger to give sharp transitions. The reasons for the asymmetrical output of the oscillator are described later in the section on the marker blips. The histogram display that results is particularly vivid which will indicate not only "0" and "1" states but also invalid voltage levels between these two states.

As the 'scope's spot begins its travel at the left of the screen, Channel 1 (input "0" of the 74151) is addressed and the spot will indicate either a nominal 0V or 5V. After it has traced out a line of one eighth of the display length, Channel 2 is addressed and the spot will either continue along the same line or switch to indicate the other value. This will continue until all eight channels have been displayed. As Channel 1 is about to be addressed again the syncronising pulse causes the spot to return to the left-hand side.

Provided that this happens fast enough a histogram display will result (although without marker blips). The rate of addressing is dependent entirely on the frequency of the oscillator. This is not critical but if the rate is too slow a disturbing flicker will mar the display: if too high, the quality of the image deteriorates due to pulse rounding and distortion due to stray capacitance in the connecting leads. While the display could be sharpened using a Schmitt trigger at the input to the 'scope or by using low capacitance connecting

leads, there is really nothing to be gained by a very fast addressing rate. Bear in mind that harmonics of the order of twenty to thirty times the oscillator frequency are present on the output of the unit as it is giving a square wave output.

It was found that an oscillator frequency of the order of 1.5kHz gave a clean flicker free display using two feet of connecting lead from the board to the 'scope.

The lead used was ordinary connecting wire, the data, sync and earth lines being just simply twisted together.

SYNCHRONISATION

Obviously if the oscilloscope used the data stream from the board to operate its internal synchronisation circuits, the display would lock into a random position. A convenient source of a synchronising pulse is available from pin 8 of the counter. This goes negative as Channel 1 is about to be addressed. Provided the 'scope can be set to lock on a negative edge, Channel 1 will be on the extreme left.

If it is preferred to lock on a positive pulse the sync output can be inverted or the inputs renumbered so that what would have been Channel 5 is numbered Channel 1, Channel 6,

Channel 2, etc.

COMPONENTS... Resistors R1 620Ω R7 10kΩ R2 $10k\Omega$ R8 360Ω R3 $10k\Omega$ 620Ω 360Ω R5 R6 10kΩ Capacitors 0.047µF C1 C2 0.1µF C3, C4 10µF (2 off) 25V elect. C5 0-14F Integrated Circuits IC1 7413 IC2 7490 74151 IC3

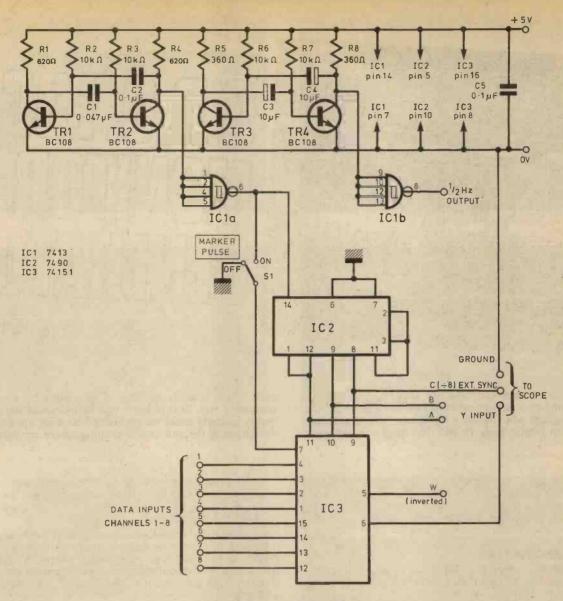


Fig. 1. Circuit of Logic Monitor

OPERATION

There are three 'scope connections: data output, sync output and common earth.

The 'scope should be set to give a reasonable size of display using an input sensitive down to d.c. as otherwise the display will keep moving around the screen. Unconnected inputs will give a "1" display so ground one input, say Channel 2 and adjust the 'scope timebase until the image locks. If the time base is running too slow more than one image of the display will appear side by side.

Increase the timebase rate until a single image appears. If the timebase is too fast the image will fold over on itself. Experiment with grounding various combinations to check that each input has been connected correctly.

A length of 10 way colour coded ribbon cable can be used, allocating colours to the channels following the resistor code. (Channel 1=brown, Channel 2=red, Channel 3=orange, etc).

The black and white wires are used to carry the 0V and 5V supply from the circuit under test to power the board. The connecting leads can be terminated in any convenient way.

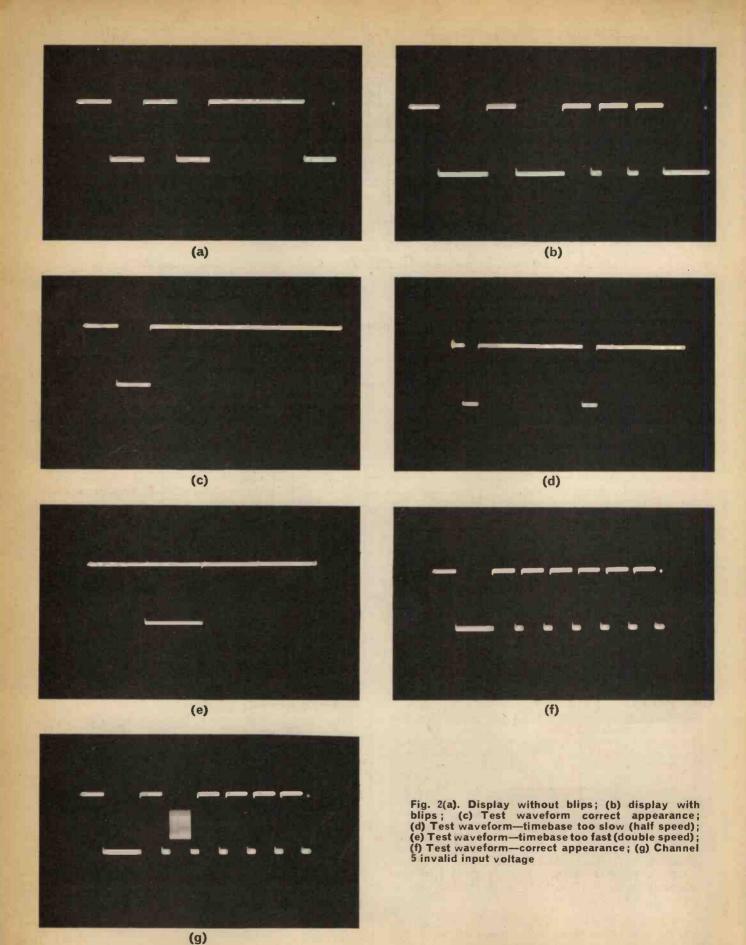
MARKER PULSES

As described so far the circuit is fairly conventional and gives a straight histogram display so that if all the inputs are at "1" or "0" a straight line results which can be less than clear as a display and some form of marker blip is desirable.

This can be achieved without additional components in two ways. The first makes use of the strobe input of the 74151. Unless this is grounded its output stays at 0V. If this input is fed with the output of the oscillator the output is returned to "0" for the duration of the positive half cycle and follows the condition of the selected input for the duration of the negative half cycle. This is why the oscillator is given an asymmetrical output.

By making the positive excursion about one sixth the length of the negative excursion, small blips are created on the zero line.

If you have an oscilloscope with a sensitive "Z-axis" modulation input the asymmetrical oscillator can be fed to this instead. The display will be blanked between each section of the display, giving discrete dashes.



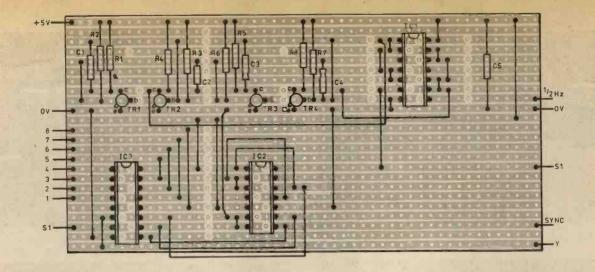


Fig. 3. Veroboard layout

PERMANENT INSTALLATION

If desired, the board can be permanently mounted in a box on the workbench. In this case it is convenient to provide a "Marker Blip" on/off switch.

This is a simple changeover switch that connects either an earth or the oscillator to the strobe input. A second switch could be added to switch the 'scope away from the logic monitor circuits to look at logic waveforms directly—care being taken to keep stray capacitance to a minimum. An additional convenience would be switches that would disconnect individual input channels and substitute an earth so that unconnected inputs will indicate "0".

It was found convenient to mount a 0.5Hz oscillator on the same board. This is useful for cycling circuits under test slowly for examination of their working.

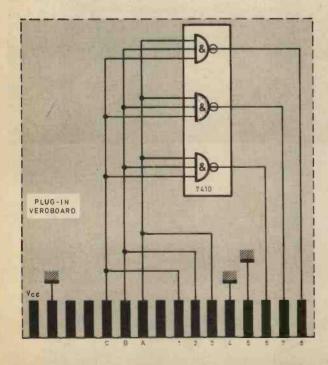


Fig. 4. Typical test set-up

TESTING INTEGRATED CIRCUITS

The testing of TTL i.c.s and i.c.s that are feared to have been damaged is speeded using the display. Using a \$\mu\$Dec breadboard, a 7493 divide by sixteen counter is used to provide inputs to the gates under test. Both inputs and outputs are monitored. For example, to test a 7410 triple 3-input NAND gate i.c. the A, B and C outputs of the counter are connected to the three inputs of each gate and also Channels 1, 2 and 3 of the display. The gates can now be seen cycled through all eight possible input conditions as the counter is pulsed by the l.f. oscillator. Channels 4 and 5 are grounded for clarity of display and Channels 6, 7 and 8 connected to one each of the three outputs.

With a correctly operating i.c. all three outputs will go to zero simultaneously when three "1s" are at the inputs of the gates. A faulty gate will be seen to be behaving differently. By examining the display as the circuits are cycled it can soon be seen if the faulty gate can be used as a 2-input gate or as an inverter. Once the breadboard has been set up it is a quick and simple matter to unplug one i.c. and substitute another of the same kind.

If you regularly use "untested" i.c.s it might well be worth your while to wire up a series of plug-in pieces of Veroboard that would instantly set up the correct conditions of a range of commonly used i.c.s. Some of these boards might need subsidiary logic on the plug-in board but the necessary i.c. need only be plugged into a socket and can be borrowed from another logic circuit not in use at the time of testing. (see Fig. 4)

INCREASING THE CHANNELS

The number of channels can be increased to sixteen by replacing the 74151 with a 74150 sixteen to one line multiplexer driven by a 7493 divide by sixteen counter. A lesser number of inputs can be arranged by resetting the counter before it reaches sixteen.

The 74150 gives an inverted output and will give an "upside-down" display unless its output is fed through an inverter.

Theoretically the number of channels can be increased indefinitely by using additional data delectors and counters combined with additional analogue circuitry to give pseudo double, triple or quadruple beam displays. However, the more channels that are used the higher are the harmonic frequencies that have to be handled with the resulting difficulties of keeping a good pulse shape on the 'scope along with the screen becoming cluttered with too much data. In practice eight channels are more than adequate.

DIMWAT



M.J. CARTER

A sympathetic bedside light!

NE OF my vices is that I read in bed, and tend to fall asleep with the bedside light still on. Those similarly afflicted will confirm what this can mean in disturbed sleep, not to mention electricity costs. Moreover, those attempting to help me, by turning off the light when I am found asleep, only succeed in waking me up again; as much by the transition from light to dark as by any noise they might make.

APPLICATIONS

In its original role of "humane bedside light", the prototype Dimwit has given many months of appreciated service. It has also been found suitable for ordinary lamp-dimmer applications, and to have "entertainment value" when soft lighting is called for—with an inbuilt practical joke. Other applications include porch and corridor timed lights, and nightlights for reluctant progeny.

CIRCUIT DESCRIPTION

The lamp is controlled by a mains triac circuit based on Fig. 13 from "Power Control With Triacs" by W. Williamson, P.E., March 1971. Automatic control is achieved by putting a light dependent resistor in series with the potentiometer, and illuminating it by an l.e.d. This home-made optoisolator is very much slower than the packaged sort, and requires a lightproof container, but its output is more amenable to this application.

The sequence is started by pressing the Go button S1 in Fig. 1, which applies mains power to the PSU via S1b and starts the Timer via S1a. The Le.d. is then turned on, which causes the load lamp to be dimmed up. The relay RLA is also energised, thus bypassing S1b so that power is maintained when S1 is released. At the end of the timed period, D1 and hence the lamp, will dim out slowly. Then the current in RLA coil is also reduced until it drops out, which switches off the whole circuit. The sequence of events may be followed in more detail in Fig. 2. In addition, the timer is arranged to restart if S1 is pressed during the timing period, and to "time out" prematurely if the KILL button S2 is pressed.

Referring to the circuit diagram of Fig. 3, the timer is a standard 555 arranged as a monostable, whose timing is set by R1 and C1; the value of R1 was chosen empirically to give a period of about fifteen minutes. The values used give a theoretical time of:

T 1.1 C1 R1 $453.75s \Rightarrow 7!$ minutes.

There are three reasons why the time achieved in practice is twice the calculated time.

Firstly, the tolerance of high value electrolytic capacitors (typically -10°_{\circ} , -50°_{\circ}) often make the nominal value an underestimation.

Secondly, the threshold current drawn by IC1 (pin 6) shunts some of the charging current in C1, and this stretches the timing period. Using a low leakage tantalum capacitor, this effect was found to extend the period by some 20° , over the calculated time.

Thirdly, charging current in C1 is reduced by its own leakage current, which, together with the threshold current, extends the period by just over 80° o. Unfortunately, the leakage current of aluminium electrolytics tends to increase with storage time, due to deforming of the insulating layer in the absence of a polarising voltage. Although this leakage is small, it is enough for a previously unused capacitor in the C1 position to give extra long periods for the first few timings (up to an hour or more), and in some cases preventing triggering altogether. Before soldering C1 into place, therefore, it would be advisable to re-form it as completely as possible, by, for example, leaving it connected (the right way round!) across a PP3 for a day or two. An extended series of measurements with the prototype Dimwit showed that, once the leakage current had settled down, the period varied by no more than twenty seconds in sixteen minutes, which is well within the required accuracy. For those who wish to avoid this "running in" by using a tantalum timing capacitor, it should be noted that CI would need to be increased to about 200µF to compensate the absence of

SPECIFICATION

(i) When started manually, by pressing the Go button, the bedside light comes on, and remains on for 15 minutes.

(ii) At the end of this period, the light is dimmed slowly to extinction, and the unit switches itself off.

(iii) When the sequence is started, the lamp "dims" up quickly (rather than be switched straight on), which reduces sudden dazzle, and increases the life of the lamp.

(iv) The maximum brightness of the lamp, during its "on" period, is under manual control.

(v) Pressing the GO button re-starts the sequence. whenever it may be pressed.

(vi) Pressing the KILL button prematurely ends the timing period in (i), and the unit proceeds to shut down as in (ii).







leakage-current. Increasing R1 is not advisable.

To trigger the 555 Timer, a value of $1k\Omega$ was chosen for R2, giving a source resistance to S1a which is low enough to be immune to the triac switch-on spike. The more conventional value of around $12k\Omega$ proved troublesome.

Diode D1 has been added to fulfil specification (v), and if S1 is pressed before IC1 "times out", C1 is discharged by D1 and S1a, thus restarting timing. Normally D1 is reverse biased, and takes no part in circuit action.

When IC1 turns on at the beginning of the sequence see Fig. 2 (a), C3 charges rapidly via D2, D3, and D4, then completes its charging via R4, see Fig. 2 (b). This is transmitted through the buffer stage TR1, TR2 to D1, which dims up rapidly but not instantaneously, see Fig. 2 (c). This is in order to fulfil specification (iii), and the resistance of R13 then falls. Fig. 2 (d) shows the inverse of this, so that the triac control circuit dims up the load lamp rapidly. R13, the ORP12 light dependent resistor, particularly at the low illumination levels used in this application, has a somewhat slow response, as indicated in Fig. 2 (d); the only effect this has on the completed Dimwit being however, that the mains lamp is a little more sluggish in dimming up after an hour's rest than after only a few minutes. Since the maximum peak voltage rating of the ORP12 is 110V, the potential divider

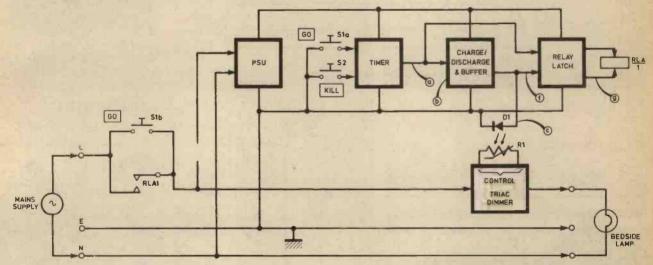
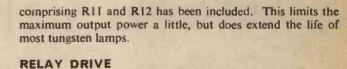


Fig. 1. Block diagram of Dimwit



The relay is energised by TR4, Fig. 2 (g), which is turned on by IC1 via R8. A handy miniature relay was used for convenience, and a possible type is the 6V miniature open relay from RS Components (part no. 349-125), for which there is enough space on the p.c.b., but which has different mounting arrangements. Since S1b carries the switch-on current surge for the lamp, which in any case is considerably reduced by the dimming-up process, see Fig. 2 (e), the relay contacts need only be rated for the continuous current of mains lampand Dimwit. Moreover, as will be seen, the current broken by the relay at the end of the cycle is smaller still. Thus, if need be, a reed relay may be used.

Once IC1 has timed out (or when the KILL button S2 is pressed), C3 starts discharging slowly via R4, Fig. 2 (a), (b); thus the load lamp, Fig. 2 (c), (d), (e), starts dimming down

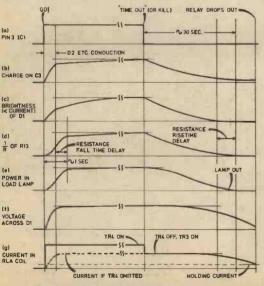


Fig. 2. Waveform timing diagram

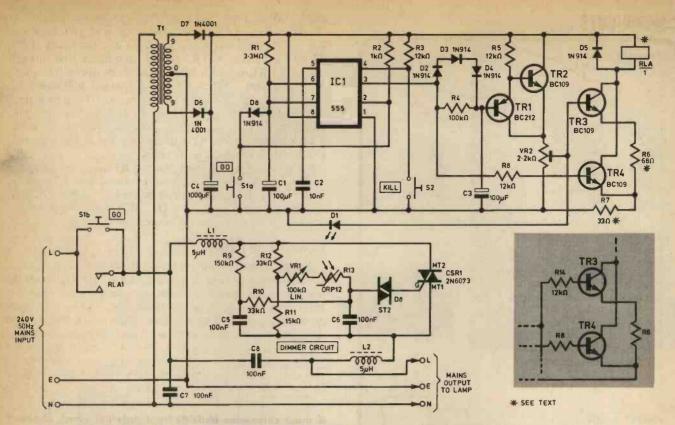
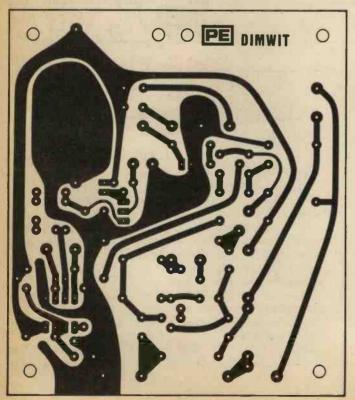


Fig. 3, Full Dimwit circuit. The shaded area shows the modification referred to in the text

Fig. 4. Printed circuit layout (actual size)



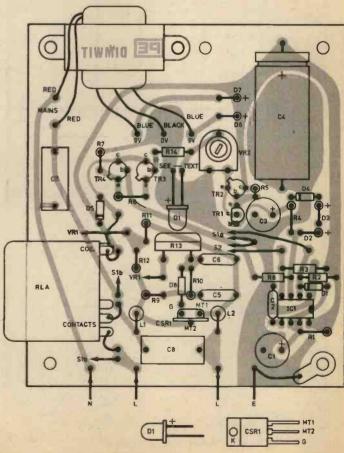


Fig. 5. Component layout

COMPONENTS ...

COMPONE	INTS
Resistors	THE RESERVE OF THE PARTY OF THE
R1	3·3M Ω
R2 R3	$1k\Omega$ $12k\Omega$
R4	100kΩ
R5 R6	12kΩ 68Ω (see text)
R7	33Ω (see text)
R8	12kΩ
R9 R10	150kΩ ½W 33kΩ ½W
R11	15kΩ ½W
R12 R13	33kΩ 1W ORP12
	5% unless otherwise stated
Potention VR1	meters 100kΩ linear
VR2	2.2kΩ min horizontal
Capacito	THE RESIDENCE OF STREET
C1	100μF/16V electrolytic
C2	10nF
C3 C4	100μF/16V electrolytic 1000μF/25V electrolytic
C5	100nF/160V
C6 C7	100nF/160V 100nF/400V
C8	100nF/400V
BURN FR	
Inductors L1	5µH v.h.f. choke, 1A (RS: 238-255)
L2	5μH v.h.f. choke, 1A (RS: 238-255)
T1	Miniature mains to 9-0-9V transformer (Altai 909 or similar)
RLA	9V min 1-pole c/over relay:
774-4	280Ω coil, mains contacts
Semicon	
D1	clear plastics red l.e.d. (5mm dia)
D2-D5	1N914 (or any surplus low-current silicon diode)
D6-D8	1N4001
D9 CSR1	Diac ST2 (or similar) Mains triac, 3A minimum (2N6073 or
	similar)
TR1	BC 212
TR2 TR3	BC 109 BC 109
TR4	BC 109
IC1	555 Timer
Switches	
S1	d.p. momentary make; 250V, ½A (RS: 337–920)
S2	s.p. momentary make, low voltage
15 coollan	THE RESERVE OF THE PERSON NAMED IN
Miscellan 1 off	aluminium box, 152 × 101 × 51mm
2 off	self-tap screws (for above)
4 off 1 off	13mm 6BA countersunk screws 6mm 6BA countersunk screw
5 off	6BA nuts
5 off	6BA locking washers
1 off 4 off	solder tag 4mm × 6BA insulating pillars
2 off	mains cable grommets
1 off 2 off	potentiometer knob 25 mm No. 8 roundhead woodscrews, with
2011	matching Rawlplugs (for wall mounting)

double-sided foam-backed adhesive tape

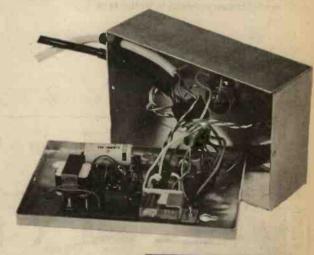
slowly, taking about half a minute to go out altogether. At the same time TR4 turns off, Fig. 2 (g), and TR3 turns on. The voltage drop across D1, Fig. 2 (f), and the values of R6, R7 set the current in TR3 emitter, and hence that through the coil of RLA, which is arranged to be above the relay's holding current. When C3 has discharged sufficiently for D1 to be off, Fig. 2 (c), the voltage across D1 begins to fall, and reduces the energising current in RLA until it drops out. This turns off the whole circuit, and thus completes the cycle. If S1 is pressed before RLA drops out, then IC1 is triggered, and the whole sequence starts again.

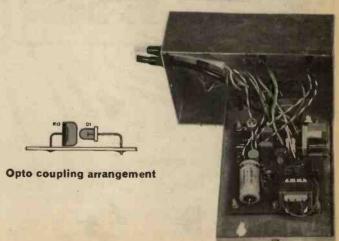
If a different relay is used, R6 and R7 may need adjusting to give correct action of RLA; if it is found that R7 must be omitted, then R14, Fig. 3, must be used to prevent TR3 base shunting all the current from D1. If R13 is not needed, its position on the p.c.b. should be occupied by a short circuit.

CONSTRUCTION

The p.c.b. and component layout are shown in Figs. 4 and 5. Construction is straightforward, as long as care is taken on the mains-voltage section to minimise the risk of flashover between tracks or to the (earthed) case. D1 and R13 are mounted as shown below. If the stiffness of the lead-out wires is not considered enough to keep them in place, they may be secured to the p.c.b. with a plastics filler. D1 should illuminate all of the active area of R13 and this may be checked visually in semi-darkness. RLA is mounted as shown below using double-sided foamback adhesive

A most chivalrous bedside light (internal view), showing the relay mounting arrangement





80mm

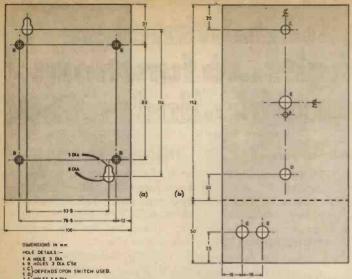


Fig. 6. Metalwork drillings. (a) Keyhole slots on backpanel, for wall mounting. (b) Front panel. Hole sizes will depend upon switch types used.

tape. The tape is wrapped around the relay case, and the relay then stuck into place. For added security, the metal clamp shown may be added. The mains transformer is also secured at one end by a p.c.b. mounting screw, and at the other with a 6BA nut and screw. Connect to mains earth.

After assembling the p.c.b., and drilling the metal box as indicated in Fig. 6, mount the circuit board on the back panel using 6BA countersunk screws and insulating pillars. Set VR2 to mid-position, and VR1 fully clockwise. Close the case but leave it unscrewed.

By connecting to a load lamp of the same wattage as is intended to be used with the unit, and connecting to the mains, full operation of the completed Dimwit may be checked. VR2 can now be set to give the desired dimming performance; turning it anticlockwise increasing the dimmingup time and decreasing maximum brightness, while shortening the dimming-out sequence. (Remember to reassemble the case after each adjustment, if a lightproof assembly is not used for R13). When adjustment is complete, screw the case together and Dimwit is now ready for use.

The completed Dimwit is intended for wall mounting. The



In the old days they managed with crude "ON/OFF" bedside lamps

"saucepan" slots, which allow the unit to be hooked onto No. 8 roundhead woodscrews, are best made by drilling a 8mm hole and a 5mm hole, and joining them with the aid of a "rat tail" file. To mount Dimwit, drill and plug the wall, to accept the woodscrews, and screw them nearly home.

CONSIDERATIONS

The "dimming up" time which occurs at the beginning of the sequence may be increased by replacing D3 and/or D4 with a $1k\Omega$ resistor, which slows the charging up of C3.

There is a faintly audible "clunk" when the relay drops out, which might be sufficiently loud to reawaken the more highly strung sleeper. If this becomes a problem, an electrically suitable type reed relay might provide an easier solution to further sound proofing, although fitting it into the available space may present problems.

It would take a number of years to recover the cost of the project, from saved electricity and enhanced bulb life, but its true value becomes apparent as you begin to take Dimwit for granted!

News Briefs

IERE CONFERENCE

THE advances made in the field of radio receivers since the days of the cat's whisker crystal detector have been considerable. It is proposed to survey the subject at a conference on Radio Receivers and Associated Systems organised by the IERE which is to be held at the University of Southampton from July 11-14 1978.

Thirty seven papers will be delivered formally and a further twenty will be presented in poster booth sessions. An exhibition of relevant equipment and systems is to be organised by the Electrical Research Association.

Further information can be obtained from The Conference Secretariat, IERE, 99 Gower Street, London WC1.

SONY BETAMAX

In April Sony announced to the press the launch of its Betamax home video recorder. With over 500,000 people using Betamax on N.T.S.C. systems in the U.S.A. and Japan Sony believe the time is right to offer a PAL standard Betamax to Europe.

Mr Akio Morita, chairman and co-founder of Sony Corporation, heralded the coming of a "Video Age". He said that Sony were not looking to capture the home video market in a rush of sales. The Sony policy is to establish highly qualified outlets for Betamax and for the after sales service to be faultless.

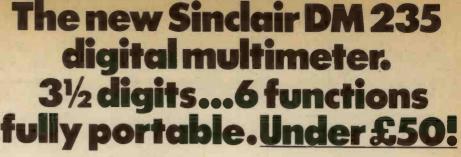
On International Standards Mr Morita said he obviously hoped that video cassette manufacturers would together establish a common system, but the inference is that Sony think their system is the best and others may follow.

Costly and specialised machinery is used to manufacture the slanted azimuth video head. This equipment is in Japan and production is likely to be based in Japan for some time to come. Sony have a TV production line at Bridgend in Wales and although it is possible that assembly of some Betamax may happen at some time in the future it must be little comfort to the Mullard TV tube work force.

The new Sinclair DM235 is yet another example of outstanding Sinclair value-engineering. Developed from the Sinclair DM2 and world-beating PDM35 (already outselling all other digitial multimeters), the DM235 provides full facilities for every application, including field servicing, testing and laboratory work. At a price no comparable digital meter can approach.

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7413	.36	7479	.85	4023	.19	4516	1.09
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Digital Thermometer

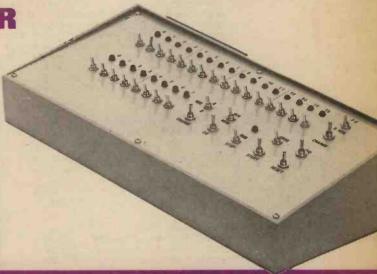
Achieves a measurement range of $-99\cdot9^{\circ}$ C to $+99\cdot9^{\circ}$ C with a three digit seven segment display. Previously, analogue-to-digital conversion involved many discrete and integrated circuit devices. This unit dramatically reduces this problem by using a Siliconix LSI chip plus a few additional components.

CONSTANT DISPLAY FREQUENCY COUNTER

AMI-COS MICROPROCESSOR KIT REVIEW

The home computer industry which has recently been booming has produced a bewildering number of "systems" for the amateur. We look at one particular kit in this review.

A straightforward unit to give frequency count capability in the range 1 to 99.999kHz with a 1Hz sampling rate. The readout does not count visibly or flicker due to display blanking. All components are readily available, and this would be an ideal project for the beginner.



ELECTRONICS

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HOW TO ENTER

The contest is for practical items incorporating electronics as a major part of the overall system. There is no restriction on the application, e.g. it could be a video tape recorder or a door bell. It may be for domestic use or for Industry.

Entries must be written/drawn clearly on one side of plain paper with the entrant's full name and address at the top of every sheet. Each entry to comprise:

- (a) a brief summary of the design (about 50 words);
- (b) any such further descriptions, drawings, sketches, photographs and circuit diagrams, etc., you consider the judges may need to form the best appraisal of your design. DO NOT SEND ACTUAL MODELS.

Each entry must have a properly completed entry coupon cut from P.E. firmly affixed to the back of the summary. Entrants may be requested to supply prototypes for evaluation at later stages of the judging. All entries must be in English.

Readers may be assured by P.E. that ideas submitted will not be misused or transmitted to other parties by anyone concerned with the competition.

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original idea not copied from any other source; (c) This idea has not been published or offered for
publication elsewhere. I/We agree to abide by the rules and conditions.
SIGNED



The closing date is August 31st 1978.

RULES AND CONDITIONS

There is no entry fee nor limit to the number of entries a reader may submit but each entry must be accompanied by a proper printed entry coupon, cut from PRACTICAL ELECTRONICS, and must bear the entrant's own full name and address. Entries will also be accepted from groups or companies, in which case the entry must be submitted and represented by an individual and the other involvements declared on a separate sheet of paper attached to the back of the summary with the

All accepted entries will be examined by a panel of expert judges including the Editor of Practical Electronics, and assessed on (a) originality of the idea, (b) technical merit, (c) practicability, (d) economic viability, (e) market potential, (not necessarily in that order). The prizes will be awarded for the best entries in order of merit. In the event of the same idea being submitted by two or more entrants, presentation of the entry (clarity, best expression, etc.) will decide such winner(s) or winning

In the event that the judges consider there are not enough entries of a sufficiently high standard, the Editor reserves the right not to award any prize(s) at his discretion.

Entries arriving after closing date will not be considered, nor will any received that are illegible, not wholly understandable, are not accompanied by a properly completed entry coupon or in any other way do not comply

exactly with the instructions and rules.

No responsibility can be accepted for entries lost or delayed in the post or otherwise; proof of posting will not be accepted as proof of receipt. No entries can be returned.

The competition sponsors reserve the right to adapt or amend any entry—after judging has been completed—for purposes of publication and/or commercial development. They also reserve the right to consider any other entries for commercial production. Where any idea is adapted or developed for commercial production payment and/or royalties and/or direct involvement in the company concerned will automatically be negotiated with the designer or person named on the entry coupon.

Winning entries will not necessarily be developed commercially, but Practical Electronics will pay the usual reproduction fee for any entries

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the competition, will be final and legally binding. No correspondence will be entered into nor interviews granted.

Winners will be notified by post and brief details of winning entries published later in Practical Electronics.

The contest is open to all readers, but those outside the U.K. may be requested to provide a British address to which any prize may be sent.

Development of any idea must take place within the U.K. Employees and the familles of employees of IPC Magazines Ltd., and the printers of Practical Electronics and anyone directly connected with the competition are not eligible to enter.

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News Briefs

MICRO-MUSCLE THROUGH COMPILERS

As microprocessor based systems are doubtless going to be designed into ever more complex systems as time goes by, Zilog, with the introduction of high level Cobol and Fortran compilers, plan to have six languages for all of its microprocessors, including the familiar Z80. Business and scientific versions of Basic, and two members of Zilog's own high level family of systems orientated languages (called PLZ), have recently been announced. High level languages, it is felt, will be absolutely essential in the next few years. "If you don't offer them", declared Ralph K. Ungerman, who is the company's executive vice-president, "you'll be locked out of the market!"

Since software generally amounts to most of the cost in microcomputer systems, Zilog have established the PLZ family to meet three basic criteria: Although high level, the language has to be able to get into the architecture of the machine and deal with interrupts. It also has to generate efficient codes because of a microcomputer's limited address capability. And finally, the compiler must be simple to implement, since a micro system has limited memory.

The full power of the Z80 is made available to you using the PLZ/ASM assembly language, and the common syntax among the PLZ programs should be easy for someone familiar

with any Algol-like language.

LASER REGORDER

For space photography, RCA has provided NASA with a laser beam recorder which generates images from all of the U.S. space agency's satellite sensors, resulting in higher quality and faster production of photographs.

Called the Operational Laser Beam Recorder (OLBR), it will be the primary photo production system at NASA's Goddard Space Flight Center. The OLBR can generate images requiring up to 20,000 picture elements per line.

NASA also has installed an RCA computerised preprocessing system to enhance the quality of the Return Beam Vidicon (RBV) camera signals from Landsat satellites which photograph the earth's resources. The OLBR and the preprocessing system allow rapid production of RBV colour composite photos that are of quality previously considered unattainable, according to RCA officials.

The OLBR is completely self-contained and self-calibrated, and is operated and maintained by a single operator. It controls, selects and accepts digital image data from high-density tape and transforms it into high resolution, geometrically and radiometrically correct, annotated latent images on 214mm strip film. The film then is developed and processed at Goddard to produce high resolution transparencies which are distributed to Landsat data subscribers throughout the world.

The film recorder subsystem accepts the timed and formatted digital words representing the picture elements and performs digital to analogue conversion and video signal processing. The processed video drives the modulation and scanning film transport subsystem which exposes the film with a shared spot, one scan at a time. All electro-mechanical adjustments and optical path film transport are of the highest possible precision to assure long-term repeatability and freedom from drifts which could affect image quality.

The preprocessor system is claimed to be unique because it provides both completely automatic and interactive handling of RBV data collected on tape at Landsat receiving sites in Alaska, California and at Goddard. Wideband video tapes are quickly analysed for quality with the locations of the

images automatically noted and indexed. The output of the system is a high-density digital tape and a computer tape summary indicating the tape contents to the operator. The high-density digital tape is the primary input to NASA's master data processing which, in turn, drives the OLBR.

ACCURACY ON THE CARDS

THINK of cards. Postcards, bingo cards, computer cards, carton blanks, book covers, in fact units of anything thicker than paper, and that could include labels, plastics sheet, film, you name it. Now supposing you need to count large quantities of these, in stacks. Counting by weight or measurement is quick, maybe, but inaccurate. By hand? Not even quick! If you're in business you might give 5 per cent over the top to be on the safe side. This eats into margins; but short measure loses customers.

You could go in for complicated machinery which separates the stack one at a time, and then counts, or you could go in for a static opto-electronic counting system. Such a machine exists, called the Optomat 520 Card Counter made by Vacuumatic Ltd., of Harwich, Essex. A beam of light is projected onto the edge of the stack to be counted, and sensing of the Lambertian reflected light to activate multiple photocells is utilised to differentiate between the light and dark transitions, caused by individual sheets and spaces. These spaces, by the way, are encouraged by the application of a compressed air stream. The photocell outputs are amplified and filtered to obtain distinct steps which then drive an electronic counter. So easy!

One of the sheets is taken at the outset, and put into a gauge so that the thickness dial can be set to control the optical scanner. A twelve inch stack is then placed against the backplate, and with the press of a foot pedal and a two second wait, hey presto, the total count appears. It is claimed that a count of 25,000 can be made in one minute with 100 per cent accuracy, and the machine will work over a caliper range of 0.005 inches to 0.02 inches at the same speed. So, humans and weighing machines collect your cards.

BLIND AUTOMATION

Most human beings look forward to the sun coming out, but a new compact control system, the Junior 10, has been developed for use with sun blinds mounted on the outside of buildings by Solar Protection Equipment Ltd., a Hampshire-based company which specialises in the design and manufacture of equipment and systems to protect buildings from the effects of the sun. The new system was field tested in a range of applications where accurate, reliable automatic control of blinds, awnings and solar screens is needed.

The package consists of a light cell, anemometer, control unit, and features include a light level selector, a wind override, and adjustable delays to control the "in" and "out" response so that instant reaction to short duration cloud patterns is avoided. There are l.e.d.s to indicate when the selected wind and light levels have been reached.

The light intensity range is approximately 3,000 to 30,000 lux, and wind speed range from 2 to 40m/sec (Force 2 to 12). Time delays on the solar bands may be set from 30 seconds to 10 minutes for the "out" mode, and between 1 and 20 minutes for the "in" position. The response delay from when the wind threshold value is exceeded to activation of the raise mechanism is permanently set at 15 seconds. Actuation of the downward movement corresponds to the delay selected on the "out" mode of the solar band. There is a manual override facility for the solar function of the unit, but which, for safety reasons, does not by-pass anemometer-controlled functions.

The Junior 10 will provide automatic operation for any number of blinds or screens on a single façade of a building. Other Solar Protection Equipment units are available to automate blinds fitted on two or more façades. The system can be used with any type of existing or new installation where the blinds are powered by electric motors.

IMPROVING BRITISH STANDARDS

Bst has published two further parts of BS 5428 Methods for specifying and measuring the characteristics of sound system equipment, a series of specifications covering the various major elements of audio equipment.

The first of the new publications is Part 3 Microphones which specifies the characteristics and methods of measurement for microphones used both in professional and domestic applications, notably wide band microphones, close-talking

microphones and those used for normal speech.

Similar data is given in the second document, Part 11 Loudspeakers, which relates to systems comprising one or more loudspeaker elements and such relevant devices as built-in-cross-over filters, transformers and any other passive elements.

Both specifications are identical with the corresponding parts of the International Standard, IEC 268. Additional British Standards, now in course of preparation, will deal with Performance specifications for high-fidelity audio equipment.

BS 5428 Part 3 (Price £7.80) and Part 11 (Price £5.60), may be obtained from BSI Sales Department, 101 Pentonville Road, London N1 9ND.

IMPROVING WORLD STANDARDS

STAND to attention, you are about to be standardised still further, this time by the International Electrotechnical Commission (IEC), the organisation responsible for the preparation of worldwide standards in the electrical and electronic fields. They have just issued IEC Publication 268, Part Nine—Sound System Equipment: Artificial Reverberation, Time Delay and Frequency Shift Equipment—applicable to equipment commonly used for obtaining special effects on sound recording, broadcasting and public-address systems.

The latest standard now provides engineers with the list of characteristics to be specified for artificial reverberation equipment, broadly described as amplifiers designed for special purposes. In addition this publication includes not only reverberation times, frequency responses, ambient acoustical noise levels and echo and delay times, but also describes relevant methods of measurements. IEC standards for other amplifier characteristics are outlined in IEC Publication 268: Sound System Equipment—Part Three: Sound System Amplifiers.

readout

... a selection from our postbag

Readers requiring a reply to any letter must include a stamped addressed envelope. We regret that we cannot answer any technical queries on the telephone.

Word Cost

Sir—It seems to be a common misconception that the cost of a microprocessor increases with the size of the word.

The reverse is probably true if one's aim is to run useful programs rather

than merely play with a toy.

A minimal useful system is a Central Processor Unit (CPU) plus a hex keyboard and a 4 digit hex display. For 4, 8 or 16 bit systems the cost difference is not likely to be very great. The real expense comes in buying a memory of useful size. For a given end result this last cost is likely to be least with the 16 bit system.

Program storage is approximately the same, at about 2 bytes for each instruction, for the 3 word sizes, but increased word size gives vast improvements in program efficiency. A single 2 byte instruction in the 16 bit system may equal as many as 20 instructions in 4 bits and somewhere in between for 8 bits. Programming is thus much simpler with less chance of errors and the saving in memory is immense.

Evaluation studies have shown that the National Pace 16 bit CPU offers the most to the beginner with aims for the future. It costs under £20, can work with both 8 and 16 bit data and, if necessary, will address up to 128 kilobytes of memory. One does not have to buy it all at once, as some writers have suggested with other CPUs, but can start with as little as 256 words, at about £10, and add as required

to any limit. A special feature of the Pace chip is the instruction set which seems especially designed for easy programming with hex machine code, permitting easy relocation of sub routines even in complex programs. It certainly seems easier to program than the various, much lauded. 8 bit CPUs.

R. G. Silson, Tring, Herts.

Any readers with differing ideas?—Ed.

Cover Blown

Sir—As professional engineers of some standing, we are, to put it mildly, somewhat amused to see the May cover of P.E. The advertised article is for a TTL logic tester, yet on the cover photograph the tester is shown in "action" on a circuit containing cmos logic.

This anomaly unfortunately confirms a few suspicions held by the writers regarding the integrated circuit industry; namely that cmos is really TIL with a self-destruct facility.

Yours explosively, Messrs. Maclean and Stanbury, Brentwood.

Well done lads. You may endorse our photographic licence as it was not thought that the numbering on the CMOS devices would show up so clearly. Full marks for our photographer.

Readers familiar with the Linear Capacitance Meter p.c.b. may have realised that the test clip was held over a blank area of board.—Ed.

TV Radiation

Sir—I note with interest the comments made in the May issue of Practical Electronics regarding the difficulties of establishing a Citizens Band service in the U.K.

Previously, I was indifferent to the proposal, but after reading the point raised in Industry Hotebook, I am now

wholeheartedly in favour!

It was stated that T.V. viewers in the U.S.A. were being subjected to interference from CB sets, and that Texas Instruments, in conjunction with the FCC were developing a high performance T.V. receiver less susceptible to such interference.

Speaking as a short wave listener, I would welcome any proposal to reduce interference from T.V. sets, and clear the airwayes of the pollution they produce. Everywhere you look, it is emphasised that no disruption of T.V. viewing is to be caused by any electrical or electronic equipment, but the problem of radiation from T.V. receivers is never mentioned. It is the age-old dual standard rearing its ugly head again, one law for the powerful, and another for the weak. To be blunt, I would rather allow C.B.'ers a couple of MHz airspace than be deprived, as I am at present, of five MHz which is totally submerged beneath a sea of T.V. radiation. and I am sure there are thousands like me.

> Steve Price, Doncaster, Yorks.

Tomorrow's Broadcasting

THE TECHNICAL POSSIBILITIES

DR. BORIS TOWNSEND, Head of the IBA's Engineering Information Service, gave the first of this year's IBA Lectures at the Authority's headquarters in February of this year; this feature is based on his lecture.

NE sees clearly the road a few yards ahead, makes out the shape of the cross-roads shortly to be reached, and even discerns the outlines of the hills in the distance. So whereabouts on our journey through time are we now? From where does our future begin?

We nightly bring into the majority of our homes pictures and sound which are beautiful to the eye and a delight to the ear, and often to the intellect, at a price which most of us could once afford. I think our broadcasting is great! We roam the earth at will—not at our will, of course, but at the will of the programme director.

But it has to be said categorically that, in the same way as America has the largest backlog of antiquated technology of any country in history, so we are stuck with old-fashioned television transmissions which are a hybrid rag-bag of improvisation and almost incompatible technologies, which carry a quite unnecessary amount of repetitive data yet still reproduce inadequate detail, inadequate contrast, inadequate colour gamut, and pictures which lack most of the attributes of real life. It is a system which gives the viewer and listener little choice and to which any graduate tea-boy engineer could now suggest improvement.

What are we going to do about it? In the next few years—not much. We did something about our equally out-moded radio transmissions, and duplicated our use of precious wavelengths with our stero f.m. transmissions using circular polarisation. These transmissions are glorious. What happened? Nothing! The public ignored them. Everyone still listens on antiquated amplitude modulation receivers—statistically, so to speak. But engineers are not easily discouraged.

SMALL IS BEAUTIFUL-AND CHEAP

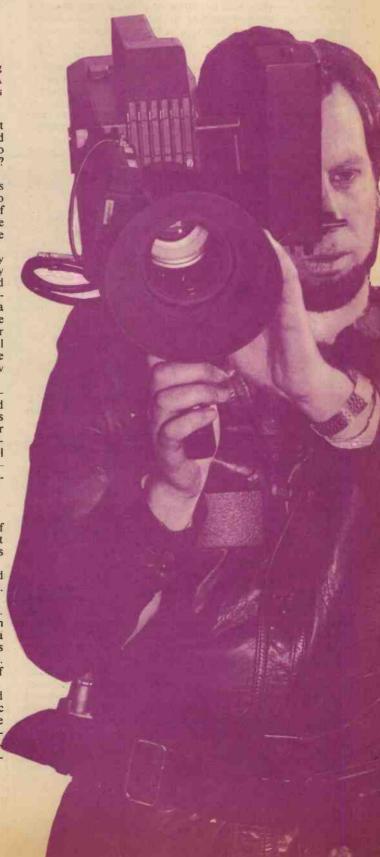
As an expatriate Londoner perhaps you will forgive me if I quote my favourite piece of London graffito: "God is not dead—he is alive and well and working on a less ambitious project!" The same goes for engineers.

The other end of the television system from the home and

The other end of the television system from the home and hearth is, to engineers, more amenable to radical change. And radical the change is going to be.

Engineers are making a two-pronged thrust forward. One is a miniaturisation of the analogue techniques which have been used in our studios since Savoy Place and Alexandra Palace, while the other is based not on smooth, continuous signal waveforms but on abrupt discontinuities, or digits. Both advances are dependent on the new technology of micro-electronics.

It is not possible to live in a developed country in 1978 and be unaware of the incredible reduction in the size of electronic circuits which has been made in the last decade. Gone are the separate valves, resistors, condensers and wires of yesteryear. Today's large-scale integrated semi-conductor micro-circuit can house thousands of individual electronic components on a small chip of silicon the size of a screw head.



With small size comes a reduction in the cost of materials, a reduction in weight, usually a reduction in power consumption and a reduction in operating time, since the signals moving through the circuits have less distance to travel. The failure rate also falls since the number of mechanical or soldered joints and the opportunities for operator mistakes are alike reduced. The cost of making an integrated circuit is proportional to its area and is more or less independent of its circuit complexity. The reduction in cost is staggering. An integrated circuit now costs about the same as an apple. As a result it is now practicable on the one hand to undertake electrical processes and functions of great complexity, and on the other hand to mass-produce cheaply for the home, devices which were once only to be found in capital-intensive studios.

Will this trend continue? It can. The latest microcircuits are approaching size-reduction limits imposed on the photoetching process by the wavelength of light—that is to say, individual parts of the circuit are not more than a few wavelengths of light in size. Further reductions require radiations of shorter wavelengths than light, and laboratories are currently utilising ultra-violet, X-rays, and electron-beam pattern-generation machines to continue this reduction in size. It is an industry which is aiming at a dimensional precision two orders better than that to which mechanical engineering machine shops can work.

MECHANISMS INTO MICROCIRCUITS

Which brings me to a point of fundamental importance in our forecasting. The more recent history of manufacturing industry is that, despite the advantages of large-scale production and the replacement of labour by automatic machines, mechanisms grow more and more expensive to produce, while electronics become cheaper and cheaper.

Compare the £70 monochrome television receiver, of dubious performance, which sold in 1938, with its contemporary £100 small motor-car: and then take today's large, bright, reliable monochrome television receiver, still at £70, and its contemporary £2,000 mini. So whenever we can replace mechanisms by electronics we shall. Whenever we can transfer a problem from the mechanical domain into the electronic domain, we shall—and indeed already do.

These tiny microcircuits already make broadcasting equipment portable in a sense little dreamt of by the early pioneers. Every swinging teenager has the ability to make an audio broadcast tape of better quality than the phone-in programmes we permit. Sophisticated editing is possible in any one's back-room, while multitrack tape-recording and its subsequent re-processing can restore a semblance of protessionalism to any amateur pop group. All this opens up new avenues of exploration to enterprising programme directors. And a parallel development is going to take place in television. Yet there are still surprising weaknesses in the equipment available for radio reporting, as will be testified by roving radio news crews attempting to establish a link back to base, or even to aim a microphone at an American President.

ENG

In television, the evolutionary reduction in size and weight of television equipment has produced a revolutionary change in the medium, for as Valéry said, rephrasing Engels—similar figures exist only in pure geometry. Small, electronic, colour-cameras of low power-consumption are now in production and are giving immediate world coverage of important news events. These Electronic News Gathering developments will spread to the field production of drama and even to light entertainment programmes. If we wanted so to do, we could now operate a television service without much in the way of television studio complexes.

An essential ingredient of this light-weight facility is a portable battery-operated colour video-tape-recorder. Now to make satisfactory replays of colour video-tape-recordings the machine has to work to a timing accuracy of around one

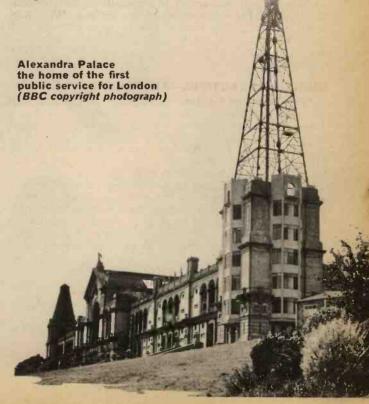
thousandth part of one millionth of a second. It is not possible to pull a magentic tape through a VTR with this precision, even on a massive and expensive studio machine and certainly not on a small portable recorder subject to the gyroscopic effects of being carried or even of being used on a motor-cycle.

The technological solution has been to transcend the problem by abandoning the attempt to make such a precision mechanism. Instead, we make a small, cheap tape-transport mechanism and then correct all the recorded and replay timing-errors electronically, and subsequently, back at base—a typical example of our foresaking mechanisms for microcircuits. The ability to carry out this electronic correction in a Time Base Corrector depends not only on a high complexity of microcircuits but also on the digital techniques of the second prong of our forward development thrust—of which more in a moment.

ENG equipment is now sufficiently automatic to be used by any sensible person—providing maintenance back-up is available. Thus, last year, when NBC was suffering a technicians' strike, the manager of one station gave two of his girl secretaries half-an-hour's instruction in the use of the ENG equipment and then sent them out to cover a story—which was later transmitted. This is not so much a cautionary tale for strikers and strike-breakers but more an encouragement to all actual and budding directors. Television programme-making is no longer necessarily capital intensive. Television directors need no longer rant about there being 20 engineers between them and their actors, while programme-makers are no longer dependent on the owners of expensive studio complexes for permission to ply their trade.

FILM

Whilst I am still talking about analogue television I should not ignore film. CBS, of course, would say "Why not?—film has gone from television!" As a forecast rather than a contemporary fact, CBS may have it right. But I cannot refrain from observing that the period of greatest innovation in the history of the sailing boat was in the 50 years following the invention of the steamship. We have not heard the last of film. I have held in the palm of one hand a working prototype of a professional 16mm film camera which is so much smaller than any 3-tube electronic camera that the uncommitted observer is bound to say, why all the fuss about lightweight ENG?



Certainly, live ENG is more immediate than film-but most ENG is not live. Film processing times are already shorter than the average time between ENG recording and ENG transmission, whilst Dr Land has demonstrated more or less instant-processing polaroid colour cine film using an additive colour process.

It may be that electronics, rather than killing off film, will save it; for as our electronic picture processing becomes more sophisticated, so we can rescue film from its own defects of colour mistracking, variable colour grading, frame bounce, spots, even grain—but we may not wish to do so.

DIGITS

But I keep straying, inevitably, in this review of where our technology is at the moment, into the second prong of our forward development thrust, which is based on computer technology. From continuously varying analogue signals in the studio and network we are about to change over to trains of constant amplitude pulses—to digital radio and digital television, using components developed for defence and for computers. Why?—you may ask. Are our pictures going to be sharper, or more colourful, or bigger, or be stereoscopic? Is our sound going to encompass a wider frequency range or have more attack? The honest answer is

Certainly digital signals will travel long distances better than analogue signals; that is to say, with less distortion and without picking up noise en route. Certainly digital—or television-by-numbers—equipment can eliminate the line-up time of studio equipment and can reduce maintenance routines; however, for most of our present-day studio processes, digits do not yet offer any cost saving. Certainly they permit us to undertake an incredible range of picture processing and correction and special effects.

DICE

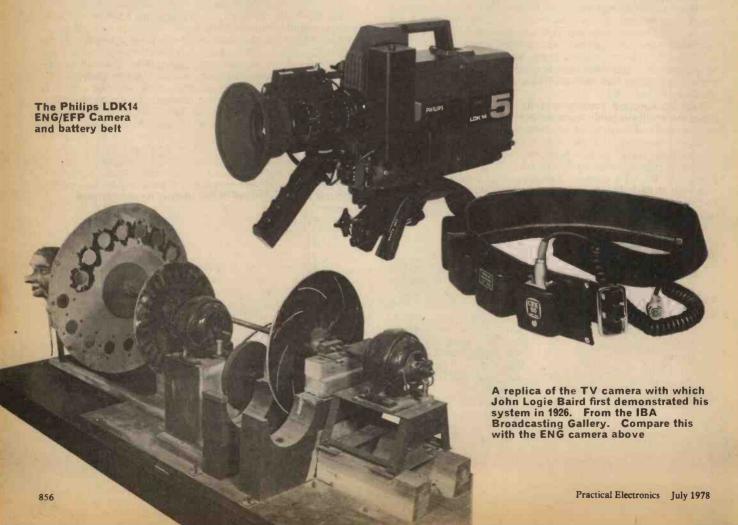
Digits enabled us to put into everyday operation, some five years ago, an all-electronic colour-picture standards-convertor to change programmes from the American System to the British System, and conversely, using electronic arithmetic. This convertor, named DICE, needed the equivalent of some five million valves, packed into a space about the size of a wardrobe. In Venice last April we demonstrated to the Technical Committee of the European Broadcasting Union most of the components of an all-digital television studio; and we are now negotiating licences for the manufacture of digital video-tape-recorders to our IBA patents.

Almost anything goes in the field of special effects, including compressing and inserting one camera picture into part of another, flipping the picture about an axis, reprocessing ENG signals to remove noise, and most of the other defects due to simple cameras, and lots of other processes which engineers had long regarded as impracticable. And, unlike film special effects, the results are available immediately.

Digital pulses also enable us to add services to our existing broadcasts. By placing a string or two of fast pulse trains in the upper blank margin of our television pictures, engineers have given us a new information service—our ORACLE Teletext service, complete with simple computer graphics as well as typefaces. It uses a great deal of computer technology both at the studio and in the home receiver. Although the receiver ORACLE circuits are much more complex than all the rest of the colour set put together, the additional cost of the ORACLE decoder is going to fall well below the cost of the colour receiver alone.

MICROPROCESSED NETWORKS

Computer technology can make decisive inroads into television operations as well as into television signals. The microprocessor, which is the central brain of computers, is now controlling many routine functions in television as well as in industry at large. It is this kind of innovative technology,



and the management which is needed with it, that enables us to operate and maintain 400 transmitters with the same number of staff as we needed for only 40 transmitters some dozen or

so years ago.

The automation of our network continues at an increasing pace. In the next few years we shall reduce the control rooms for our country-wide transmitter network from fourteen to four. Our new Regional Operations Centre, or ROC, at Croydon is controlling and monitoring some sixteen main stations with their nineteen relay transmitters stretching from the Channel Islands to the Wash.

Microprocessors measure the signals pouring out from our transmitters and then display on our central television monitors, which are now more like computer-type visual display screens, not only data on the state of our transmissions, but coloured computer graphics of the circuits of any faulty transmitter, the fault diagnosis, the action which the transmitter may have already taken to cure its own problem, or the recommended repair procedure if human intervention is called for. This is all daily, routine, run-of-the-mill operations for us. This is our IBA present, the base from which I shall be leaping into the future. Of course, it will be a radio future as well as a television future.

In radio, engineers are experimenting with surround sound systems as the next step forward from stereophonic sound—or perhaps I should describe it as a step backwards from two loudspeakers in front to two more behind you. Our ILR companies have already made several successful quadraphonic broadcasts.

NEW STUDIOS

In television studios the 2 inch quadruplex video-taperecorder is now clearly in its death throes, and will shortly be delegated to archival replay only. The 1 inch analogue helical scan recorders offer reductions in capital and running costs but will not have time to replace the 2 inch quadruplex machines before the 1 inch digital recorders get into largescale production.

Digital telecine machines with one-line readouts and simple mechanical transports with zero line-up time will appear. The studio camera pedestal can give way to the stabilised hand-held camera, and studio floors can develop ridges if they like, while developments in fade-reduction techniques for radio-microphones, together with a marked reduction in their size, will eliminate, at long last, the boom shadow.

Real background scenery will be unnecessary, although super-stars will probably insist on having it. Toy scenery and chroma-key techniques will do, with computer control of the scenery camera, which looks into the miniature stage through a periscope, keeping the perspective of the background correct as the foreground camera is crabbed. After a few years, even the toy scenery will disappear. The scenery camera will look at flatly lit photographs of the required background and the graphics computer will adjust the perspective and the shadows of the transmitted image as required.

Post-production editing will increase and clever microprocessor control systems will speed-up editing for both film and tape. The acceptance of more natural-looking lighting in television pictures will combine with continuous and simultaneous tape recording of the action as seen by each of several electronic cameras, with all cutting, mixing and editing taking place in the calm atmosphere of the post-

production suite.

The technical need for special lighting for television will disappear. Already the Lincoln Centre in New York televises its theatre productions of the Metropolitan Opera on the simple basis that the cameras must not intrude on the audience, and that the lighting and stage production will be exactly those which the stage director wants for his theatre audience; that is to say, many scenes have a huge contrast range and the mean lighting level between scenes ranges from deep gloom to high intensity. The severe technical constraints thus placed on the television crew are giving rise to novel signal-processing ideas of wide application. By 1984, cameras



will automatically produce suitable television pictures almost regardless of the scene lighting.

Microprocessors and computers can play a major role in studios as well as in networks. Several broadcasters have already built very large studio complexes in which the technical facilities are assigned between studios by computer. It is technically feasible to operate the whole television system from VTR and telecine replay, through the continuity suite and master control, to network switching and transmitter, under the immediate control of computers rather than operators. The day's transmission schedule can be compiled in the computer memory over the preceding fortnight or so, item by item, as convenient. On the day itself, the computer will give a running commentary to the presentation controller on what it is doing, and on what it is going to do, and on the optimum way of slotting-in last minute re-arrangements.

International discussions are going on about standardising the coded data which should be placed on the leader of every recorded programme to ensure correct transmission from automatic studios anywhere in the world, once the machine-minders have loaded the reel onto a replay machine—any replay machine. If the machine-minders put it on the wrong machine, the computer will merely change the machine assignments. If they don't put it on any machine, then the computer will ring alarms in the rest room as transmission time approaches, and will indicate the immediate action which it requires of the humans.

DESIGNING ENGINEERS OUT

The design engineers will engineer the operational engineers out of television. The design engineers will also engineer the scenery makers and the graphic designers out of television, with chroma-key, Magicam, electronic captions and even more sophisticated electronic graphics-generators. It will be possible to quadruple the output of a graphics designer by giving him an electronic drawing board. By moving his

electronic pen over his action area he will instruct a computer to store his design and/or display it on television monitors. The designer will be able to ask the machine to do it again in a different colour, to draw it larger or smaller, to replicate the design twenty times around the edge of the frame, or to add the company logo, or to overlay any other design which he already has in the machine store, or to show it from a different perspective, or with different lighting.

Design engineers are also designing maintenance engineers out of television. One large television equipment manufacturer with a world-wide market is considering building into its products a self-diagnostic fault-analysis routine so that if one of its video-tape machines, say in the remoter parts of the Upper Volta, is too sophisticated for the local operator to handle, then the factory computer can interrogate the Upper Volta VTR via a satellite. The factory computer will then either adjust the remote television machine or will instruct the overseas human operator on which module to replace. The faulty module will then be airmailed back to the factory for repair.

π IN THE SKY?

Satellites, of course, are going to play an increasingly important role in television, despite the fact that the geostationary orbit already has some 400 man-made objects in its rather narrow equatorial annulus. We are only two years or so away from our ENG crews being able to send their news pictures straight into the sky, to a satellite, for immediate world-wide distribution, rather than going through the slow and tedious procedures of arranging terrestrial links with their PTTs.

Networked programmes can be fed to our earthbound television and radio transmitters from a satellite in space, which could be cheaper than paying Post Office charges for terrestrial links. It is clear that direct broadcasting from satellites straight into everyone's home will become practicable in the 1980s. The service is already planned. Five television channels in the 12GHz frequency band have been allocated to each European country. The heavy-platform satellite will be launched in 1981, carrying sufficient power generation to lay down adequately strong signals in European homes, and a year later it will begin to do precisely that, with television programmes.

TELEVISION SET FOR TELEVISION?

By the 1980s we can also expect the domestic television receiver to be as much a computer visual display unit as it is a television set and television will face even more competition than does radio. Microprocessors and LSIs will turn this colour VDU into an indoor games centre, covering everything from chess to clay pigeon shooting; children could do their homework on it and perhaps hand-in a magnetic printout to their teacher next day, eliminating complaints about illegible hand-writing, and permitting machine checking of the answers. It could be that the television set will be in continual use—but rarely for television, for domestic video recorders and domestic video disc players will be commonplace. I forecast further improvements in receiver safety and reliability, and that we shall have no hesitation in leaving our home equipment running all night.

All transmissions will carry coded pulses identifying themselves and their programme classification. It will be practicable to instruct the home recorder to record automatically our own selection of programmes or any programmes bearing a specified teletext code, to be stored ready for viewing, as convenient.

Thus the recorder could play back to us over breakfast our own telenewspaper with a make-up previously decided by ourselves, such as only news items concerned with football or the international price of tin, or the Middle East situation. Our transmitters and studios could run all night churning out recorded, labelled, programmes mixed with live news reports, while our domestic recorder compiled our own minority-taste programmes. Of course, one of the ORACLE Teletext

pages will be providing captions to the programmes for the benefit of the hard of hearing.

I am saying that we could now take broadcasting out of real time, with a more convenient service to everyone, better utilisation of our capital investment in transmitters and receivers, and a better service to minority viewers and listeners. It seems to me that our development engineers may do more to save the Welsh language by keeping our transmitters on-air throughout the night, than do those who put them off-air in the evening.

WANTED

What of things which we may not have in the 1980s? At the moment I see no signs of a large, bright, practical, flat semi-conductor colour television display screen. We are waiting for the unpredictable breakthrough in material technology, which I expect to come in the middle of the next decade

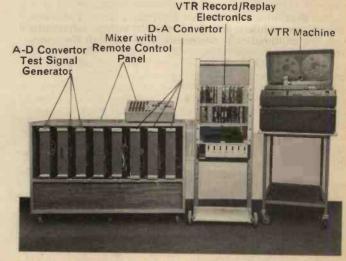
I see no indication that anyone is anxious to have 3-D television, although it is practicable with small screens if the viewer will wear polaroid spectacles, while the more dreamy-eyed physicists can visualise moving colour pictures in depth using holography.

An improved light valve, and practical methods of scanning laster beams, are still awaited; but perhaps the race for a better television display will be won by a fast-acting colour liquid-crystal display. Very fast, but dim, light modulators are already in use as sending-transducers in fibre-optic communication channels.

The cheapness of such glass telecommunication cables will make it economical to provide a wideband communication channel into every home, perhaps with a two-way interactive connection back to the local exchange. But I see no effective use of such a system until we have small, very high-density, electronic memories with multiple simultaneous access, so that central television programme libraries become practicable. Considerable research is in progress into such memories, for both defence and the computer industry, but I do not see it happening until the 1990s. When it does, we may need to redefine the terms "wireless" and "broadcasting".

RADIORACLE

Our radio transmissions will carry additional pulses to indicate on the receiver tuning display the name and frequency of the station we are tuned to—further pulses will be relayed from the radio to our visual display unit—sorry, part-time television set—to give us sleeve notes on the music being played, or the developing synopsis of the radio drama for the benefit of those listeners who missed the beginning of the story; that is to say, a form of RADIORACLE.



Experimental digital video recording systems developed by the IBA

I expect to see further development of our medium-wave radio. It is possible to extend the top note response of medium-wave radio reception in an artificial but adequate manner, and it is also possible to transmit stereo on medium-wave

It will be possible to provide out-of-city radio to motorists from a broadcasting radio satellite, but further advances in receiver aerial design are needed and will take a few years to develop. However, this could be the cheapest way of providing multi-channel radio services designed for motorists.

SMALL IS DEMOCRATIC

The television technical developments which I am forecasting carry with them their specific logic for their own exploitation, and I see two quite different forms of television service emerging.

Firstly, any local community will be able to produce its own television news and local affairs programmes, and intersperse them with old films or even dramas shot in the neighbouring National Trust Property. The small light-weight equipment, backed-up by magic black-boxes of electronics which remove all the picture defects of the tiny cameras and recorders, will be combined with cassette video recorders and video disc players to give us community television, using transmission frequencies in, or just above, our current u.h.f. bands

Such community-television studios will be modelled on our successful Independent Local Radio stations, for it will be technically feasible to operate a television company with not much more equipment than one of our present-day ILR stations. The IBA Quality Control engineer will call to carry out regular tests on the magic black-box picture processor on the sation output meets our engineering Code of Practice. The sales staff will carry ENG equipment in their cars, and between calls on advertisers will cover the local action, such as it may be. Small, in television, will be beautiful, local, democratic and very profitable.

EUROPEAN PROGRAMME COMPLEX

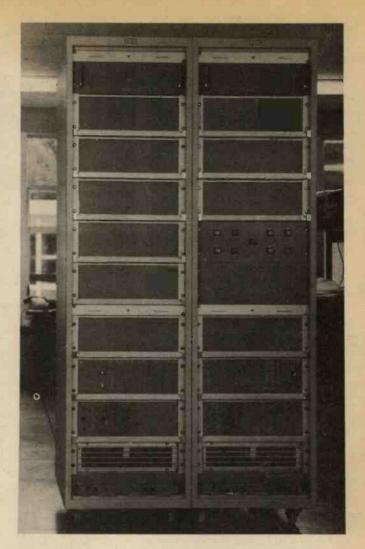
The second option which I see unfolding before us is very different, but is also based on the inexorable logic of the new technologies. A geostationary broadcasting satellite, positioned over Uganda with a reasonable northward-facing aerial, can see at least half-a-billion viewers. This size of ad-mass is worthy of a moment's thoughtful contemplation. Half a billion. There is no technical difficulty in associating numerous sound channels with each of the vision programmes, and each sound channel can be in a different language.

The size of such an audience warrants spending prodigious sums on international spectaculars, which I expect to see produced in a massive computer-run television complex equipped with every technical refinement, including security against electronic hi-jacking of the programmes. The international spectaculars would carry international advertising from the multi-national companies. It would call for entrepreneurs of stature, but then we have them in ITV.

Although I have only strayed into the middle of the next decade in my forecasting, I am already into a "Gee Whiz" situation, and I have ignored the two or three unpredictable breakthroughs in basic science and materials engineering which we can reasonably and statistically expect in this period. One could assert that there will be nothing remarkable in the 1980s, because almost anything we wish to do with our technology will be feasible. We shall have lost our ability to impress ourselves.

1984

But one of the things we may have to do by 1984 is to replan our then obsolete v.h.f. 405-line monochrome transmissions. I expect that we shall lose the use of these frequencies for television, which will be a pity, for Band III has excellent propagation characteristics for our purposes. Nevertheless, I expect the engineers to put up a case for a new television service in the band using the latest technologies—that is to



The DICE colour picture standards converter

say, they will propose a type of service which is an advance in engineering terms; for example, better definition, more scanning lines, larger pictures, tone gradation scales tailored to human physiology (and I hope a certain technical flexibility to make television more helpful in teaching the handicapped child with abnormal physiology).

1989

Our technology has reached a rather exciting stage—a very democratic stage—an open stage. Television research engineers have some powerful technologies at their disposal, and organisational methods which enable them to move fast in meeting a demand. The question is no longer "What is possible?" We have to decide between a bewildering array of possibilities. Almost anything materialistic is now technically plausible. The problem for us is of a distinctly different character from the problem facing Nipkow. And it seems to me that it is not a problem which should be left to engineers.

It is no longer a question of what can the engineers do? It is a case of what will the programme makers of the 1980s want? What will the governments of the 1980s permit? What will the public of the 1980s pay for? Who will tell us? Or is everyone saying with Thurber "Progress is all very well, but it has gone on long enough!"?

My own inclination is to tell the engineers to press on and take their own foolhardy decisions, for I believe with Francis Bacon that truth comes out of error more readily than out of confusion. But at least you should know that engineers are listening for an answer.

Semiconductor UPDATE...

FEATURING: ICH 8510/20/30 NE590/591

R.W. Coles

HORSEPOWER DRIVER

The ready availability of pre-packaged analogue circuit blocks (for example the 741 operational amplifier) has made it possible for anyone who can read a data sheet to become a designer in what used to be a difficult and complex field. The new found ability to carry out system level design has done a lot to make our hobby more rewarding, and so any extension of the "bullding-block" principle into new areas is always welcome.

The trouble is that the system design approach has its limitations. There are always problems turning up which cannot be solved with off-the-shelf integrated circuits, and at this point we can sometimes feel our design knowledge wearing

a bit thin!

Consider the problem of building an operational amplifier which can source and sink not the few milliamps of the 741, but let's say a couple of amps to drive a d.c. servo motor. You could of course hang a couple of Darlington transistors on the output of a 741 and perhaps even build in some fancy current limiting circuitry to stop the thing exploding when the motor stalls, but this would mean quite a lot of work, even if you knew how!

Well this is one particular problem which need not stump you anymore thanks to Intersil who have brought out a sort of hairy-chested 741 which, in one of its three versions, can deliver 2·7 amps with a 48 volt output swing. Coded ICH 8510/20/30, the new circuits are mounted in eight lead TO3 power transistor packages and feature full short circuit protection so that they can be used to drive up to 0·1 horsepower motors as well as solenoids, actuators and what-have-you.

Inside the can is not a single monolithic chip but a standard 741 chip married with a special driver chip on a beryllium oxide substrate to form what is termed a hybrid circuit.

DIGITALLY PROGRAMMABLE P.S.U.

One particular application example in the data sheet attracted my attention; an ICH 8530 was shown connected to a digital to analogue converter (DAC) to provide a precision variable regulated d.c. power supply which could be programmed with an eight bit word over an output voltage range of minus 25 volts to plus 25 volts with a 3 amp output current rating. Using a binary coded decimal type DAC would enable thumbwheel switches to be used to set the output voltage, and with three decades, a 0-1 volt resolution could be achieved (Fig. 1).

If power supplies and motors are not your line, well these versatile building blocks make useful audio power amplifiers too!

CHEAPER PORTS

When you start to get ambitious with your pet microprocessor, you will want to wire it up to all the switches, lamps, and lawn sprinklers which live in the outside world, and that means that you will want ports.

You will no doubt be tempted to use the all-powerful, programmable LSI parallel interface chips such as the Intel 8255A, which offers 24 I/O lines which can be programmed in groups of eight into any of three modes (Input-Output-Bidirectional).

If you need flexibility and relatively few I/O lines, this is the best choice, but if you need lots of inputs and outputs which are defined in advance, the job can be done much cheaper with standard TTL, if you know how!

TTL is best when you can justify around 64 input lines and 64 output lines as sixteen eight bit ports, and obviously not all applications need this sort of capability. For the inputs use eight SN74251 three-state-output, eight input, data selectors. Each input to a particular chip will form one line at a particular port address, the eight chips give the eight lines per port needed by most micros, and no additional address decoding will be needed.

Outputs have posed more of a problem until recently, because output information is only available on the bus for a short period, and therefore needs to be latched by the output port itself. Signetics have now introduced an ideal TTL solution to the output problem in the shape of their NE590 and NE591 eight bit addressable, latched, peripheral drivers which not only have the correct logical formulation but also the benefit of sturdy Darlington output stages which can sink (NE590) or source (NE591) up to 300 miliamps each.

The 590 comes in cheap sixteen pin d.i.l. form, like the SN74251, while the 591 uses an eighteen pin d.i.p. to allow the output stages to operate from a "higher-than Vcc" supply.

CHIP CONNECTIONS

To use these chips, connect the three least significant bits of the address bus to the address inputs and use the appropriate control signal to drive CE which latches the data into the ports. As with the input scheme, each chip latches one bit for each port, and again no extra decoders are needed.

To ensure no random states at power on time, these new devices have a built in power-on-clear, and to keep your buses healthy, the inputs are of the low power type.

You should be able to get 64 in and 64 out for less than £15; compare that to the cost of LSI chips!

20	21	22	23	24	25	26	27	ØBit	Vout
1	1	1	1	1	1	1	1	1	+25V
1	1	1	1	1	1	1	1	0	-25V
0	1	0	1	1	0	0	1	1	+15V
0	1	0	1	1	0	0	1	0	-15V
1	0	0	0	0	0	0	0	1	+0.098V
1	0	0	0	0	0	0	0	0	-0·098V
Et	c.								

The power supply can be set to $\pm 0.1 \text{V d.c.}$

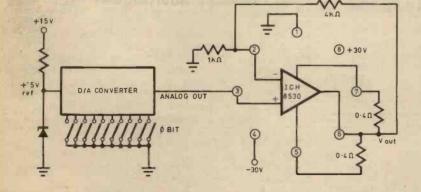


Fig. 1. A circuit which provides a precision power supply voltage. The switch program is shown in part on the right

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220V 45mA, 6·3V 2A	21 17 1
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ZA, 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60	£8-50
ZA, 6, 8, 10, 12, 16, 16, 20, 24, 30, 36, 40, 46, 60	£11.00
3A, 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60	
5A, 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60	£14·50
5, 8, 10, 16V #A £2, 12V 100mA £1, 12V 300m/	A £1.
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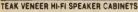
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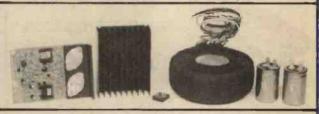
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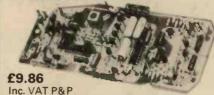
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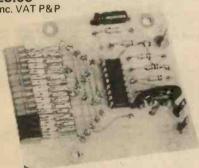
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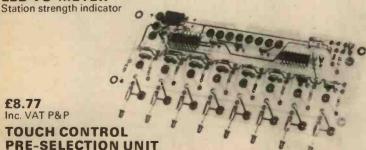
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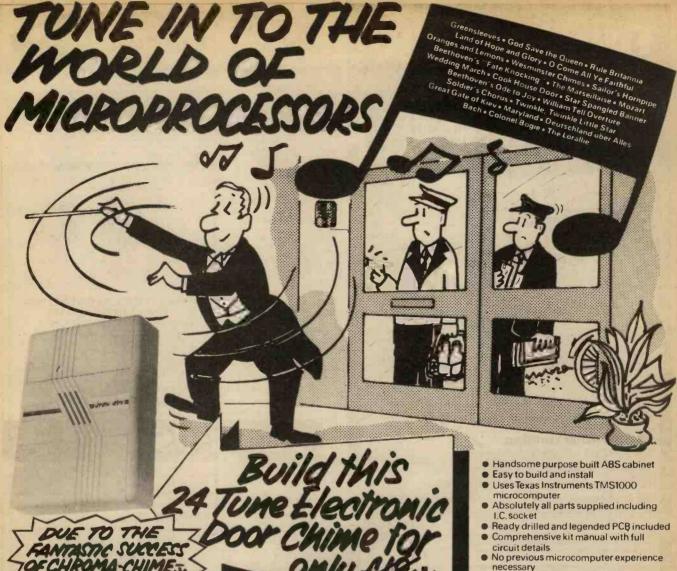


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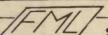
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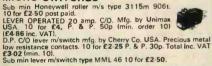
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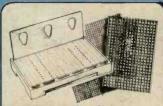
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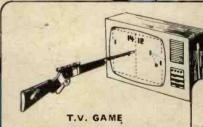
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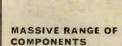
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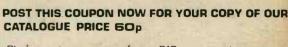
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