PRACTICAL

# ELECTRONICS

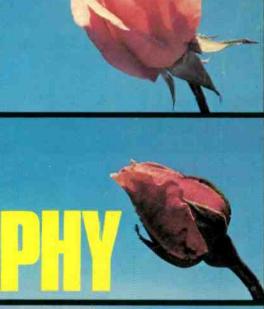
SEPTEMBER 1975

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# TIME LAPSE PHOTOGRA

ALSO IN THIS ISSUE DIGITAL I.C. TESTER



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# PRACTICAL RONIC

SEPTEMBER 1975 **VOLUME 11** No. 9

# **CONSTRUCTIONAL PROJECTS**

TIME LAPSE PHOTOGRAPHY by D. J. Bonfield Comprehensive aid to special effects photography	712
DIGITAL I.C. TESTER by W. H. Davies A quick method of testing TTL logic	72
8-CHANNEL LOGIC TRACE MULTIPLIER by A. C. Ainslie Construction and setting-up	731
P.E. JOANNA—5 by A. J. Boothman This final article deals with cabinet, board interwiring, power amplifier and fault finding	731
GENERAL FEATURES	
SEMICONDUCTOR UPDATE by D. W. Coles A review of interesting devices	728
INGENUITY UNLIMITED F.M. Centre Tuning Meter—Two Function R/C Coder	746
TRANSDUCERS—6 by P. R. Allcock Concluding article covers opto-electronic devices	748
NEWS AND COMMENT	
EDITORIAL—Private Innovators	711
SPACEWATCH by Frank W. Hyde Mars—New Comet—Venus Probe—Biological Links With Space	721
MARKET PLACE New products and a special offer to readers	737
INDUSTRY NOTEBOOK by Nexus What's happening inside industry	745
PATENTS REVIEW Thought provoking ideas on file at the British Patent Office	747
FREE ENTRY COMPETITION	
HOW INVENTIVE ARE YOU? Your chance to win a Big Cash Prize	718

Our October issue will be published on Friday, September 12, 1975 (for details of contents see page 720)

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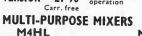
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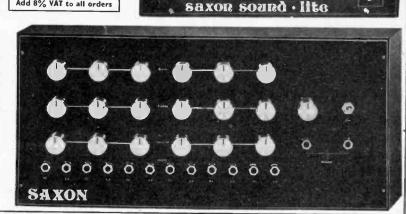
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System 1a. £65.00

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The VISCOUNT IV has a comprehensive range of controls - volume, bass, treble, balance, mono/stereo, mode selector, and scratch filter.

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SPECIFICATION: 20 watts RNS per channel 40 watts peak. Suitable 8-15 ohms speakers. Total distortion = 10 watts better than 0.2%. Six switched inputs: 1. Magnetic P.U. — 3 millivolts = 47 K ohms (R.I.A.A.); 2. Crystal/ceramic P.U. — 50 millivolts = 50 K ohms (R.I.A.A.); 3. 4. 6. Tape Tuner/Aux. — 140 millivolts = 50 K ohms (Ital frequency response); 5. Microphone — 3 millivolts = 50 K ohms (Ital frequency response).

CONTROLS: Push button ON/OFF, stereo/mono, scratch filter. 6. position rotary selector. Individual

CONTRUES: Push button Unjury, stereo/mono, scratch niter, o, position lotary selection, involved arrotary controls for treble, bass, balance and volume. Headphone socket, tape out socket. Aux, mains output. Frequency response: 25 Hz to 25 KHz # full rated output. Signal to noise ratio: better than —50 dB on all inputs. Tone control range: Bass ± 15 dB \* 50 Hz; Treble ± 12 dB \* 10 KHz. Power requirements: 200-250V AC, mains \* 60 watts. Approx. size: 15 ‡" × 3" × 10".

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de luxe plinth £20.00+£3.30 p & p. and cover Total if purchased

separately: £75.00 Available complete for only: £65.00 +£6.50 p & p.

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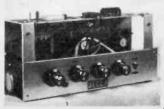
£6.50 +£1.20 p & p.

Complete with crossover Components and circuit diagram

# DECCA STEREO AMPLIFIER

Specification: 8 ohms. Input Sensitivity 4mV into 47K (for magnetic cartridges). AC Mains only 240V. Controls - volume, bass, treble, or/off, mono/stereo switch. Chassis size  $11"\times 5\frac{1}{2}"\times 3\frac{1}{4}"$ approx:

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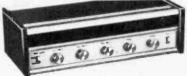
Two speakers with cabinets. Amplifier module. Ready built with control panel, speaker leads and full, easy to follow assembly instructions.

Specifications — For the technically minded: input sensitivity 600mV. Aux. input sensitivity 120mV. Power output 2.7 watts per channel. Output impedance 8—15 ohms. Stereo headphone socket with automatic speaker cutout. Provision for readphines states with administration speaker Cutous. Provision of auxiliary inputs — radio, tape, etc., and outputs for taping discs. Overall Dimensions. Speakers approx  $15\frac{1}{2}$ " $\times$ 8"  $\times$ 4". Complete deck and cover in closed position approx.  $15\frac{1}{2}$ " $\times$ 8"  $\times$ 6".

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Inputs \*4 electrically mixed inputs. \*3 individual mixing controls. \*Separate bass and treble controls common to all 4 inputs. \*Mixer employing F.E.T. (Field Effect Transistors). \*Solid State circuitry. Attractive styling.

INPUT SENSITIVITIES - Input - 1). Crystal mic. guitar or moving coil mic. 2 and 10mV. (Selector switch for desired sensitivity.) — Inputs -2), 3), 4). Medium output equipment - ceramic cartridge, tuner, tape recorder, organs, etc. - all 250mV sensitivity. AC Mains, 240V operation. Size approx: 121"×6"×31 £20.00 +£1.35 0 & 0.

# 8 TRACK HOME CARTRIDGE PLAYER



Elegant self selector push button player for use with your stereo system. Compatible with Viscount IV system, Unisound module and the Stereo 21. Technical specification Mains input, 240V. Output sensitivity 125mV Comparable unit sold elsewhere at £24.00 approx. Yours for only

£16.20 +£1.70 0 & D.



For the man who wants to design his own stereo here's your chance to start, with Unisound pre-amp, power amplifier and control panel. No soldering - just simply screw

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Also available with 2 speakers (7" x 4") £10+£1.75 p & p. f 8.95+£1.05 p & p.

# IRTABLE DISCO CONSO



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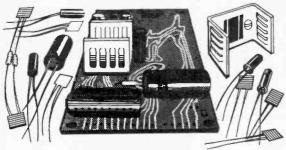
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ww	1	0 - 22-0 - 47	16	14	11 nett
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Instead of changing prices from day to day with all the difficulties that brings. EV prices are now held for 3 monthly periods and then reviewed to see if changes are necessary for the next 3 months following.

★ So far this year we have held our prices without increase up to July 1st. Next price review October 1st.

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WILLEY	Soluering in	UIIS	
C240	£2·25	Spare bits Spare bits	32p
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ole 12 way 2 pole	6 way	
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G STRIP 28 way		11p

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in lots of 100 each 4BA NUTS 34p; 6BA NUTS 28p; }\*\* 4BA Screws 30p; ½\* 6BA Screws 24p; Threaded pillars 6BA. }\*\* hexagonal £1-40; Plain spacers ∤\*\* £1-32.

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1000mF/25V 34p 2200/100 £1-91
1000mF/25V 34p 2200/100 £1-91 1000/100 £1-00 5000/100 £3-56 1000 50 50p 5000/25V 75p 2000/50 70p
5000/25V 75p 2000/50 70p
2000/25V 45p 5000/50 £1-44
POLYESTER TYPE C.280. Radial leads for P.C.B. mounting. Working voltage 250V d.c.
0.01, 0.015, 0.022, 0.033, 0.047 each 3p
0.068, 0.1, 0.15 each 4p
0-22 5p; 0-33 7p; 0-47 8p; 0-68 11p; 1-0 14p; 1-5 21p; 2-2 24p.
SILVERED MICA. Working voltage 500V d.c. Tolerance 1% or 0-5pF.
Values in pFs-2 2 to 250 in 26 stages each 6p. 270-820 in
9 stages, each 7p.
1000. 1500, 1800 9p; 2200 10p; 2700. 3600, 4700, 5000 16p; 6800, 8200, 10,000 25p.
TANTALUM BEAD
0-1, 0-22, 0-47, 1-0mF/35V, 1-5/20V each 16p
2·2/16V, 2-2/35V, 4 7/16V, 10/6-3V each 16p
4·7/35V, 10/16V, 22/6·3V each 21p 10/25V, 22/16V, 47/6·3V, 100/3V, 6·8/25V, 15/25V, each 23p.
POLYCARBONATE
Type B42540 Working Voltage—250V. Values in mF: 0-0047, 0-0068, 0-01, 0-015, 0-022 each 4p
0.018.0.027.0.033.0.039.0.047.0.068.0.082.0.1 each 5p
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9p; 0·39. 0·47 11p; 0·56, 0·68 16p.
CERAMIC PLATE
Working voltage 50V d.c.
In 26 values from 22pF to 6800pF, each 2p.

DALY ELECTROLYTIC in cans. plastic sleeved

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Double wipers for good contact and long working life
P.20 SINGLE linear 100 ohms to 4 7 megohms. each 14p P.20 SINGLE log: 4·7 Kohms to 2·2 megohms. each 14p JP.20 DUAL GANG lin. 4·7 Kohms to 2·2 megohms each 48p JP.20 DUAL GANG log. 4·7 Kohms to 2·2 megohms.
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2A DP mains switch for any of above 14p extra Skeleton Carbon Presets, Decades of 10, 22 and 47 only
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3015F Seven segment filament compatible	with standard
logic modules. 0-9 and decimal point: 9mm	characters in
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LEDS (Light Emitting Diodes)	25p
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# **BSR HI-FI AUTOCHANGER** STEREO & MONO

Plays 12°, 10° or 7° records.
Auto or Manual. A hizb
quality unit backed by SER
reliability with 12 months
guarantee. AC 200/250V.
Size 13' × 11'in.
Above motor board 3'in. Below motor board 2'in.



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PORTABLE PLAYER CABINET Modern design. Rexine covered. Large front grille. Lift-up Lid. Chrome fittings. Approx. size 17in × 15in × 7in. Few only in red and black rexine.

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MODEL P128. IDEAL AS DISCO PLAYER. 3 SPEEDS. BALANCED ARM. CUEING DEVICE STEREO CERAMIC CARTRIDGE. BARGAIN.

£15.50 PLUS POST 75p



# R.C.S. DISCO DECK SINGLE RECORD PLAYER

£6.95 Post

### SOLID MAHOGANY PLINTH Post 750

With P.V.C. Cover. Cut out for most B.S.R. or Garrard decks. Size 12½ × 14½ × 7½n.

ER 2000

# COMPLETE STEREO HI-FI SYSTEM

Two full size loudspeakers  $13^\circ_1 \times 10 \times 3^\circ_1$ in. Player unit clips to loudspeakers making it extremely compact, overall size only  $13^\circ_1 \times 10^\circ_2$  size. 3 watts per channel, plays all records 33 r.p.m. 45 r.p.m. Separate volume and tone controls.



# SPECIAL OFFER! SMITH'S CLOCKWORK **15 AMP TIME SWITCH** 0 TO 60 MINUTES

Single pole two-way Surface mounting with fixing screws. Will replace existing wall switch to give light for return home, garage, automatic anti-burglar lights, etc. Variable knob. Turn on or off at full or intermediate settings. Fully insulated. Makers' last list price £4-50. Brand new and fully garantneed. OUR PRICE £2.50 POST 350

BLANK ALUMINIUM CHASSIS. 18 s.w.g. 2½in sides 6 × 4in 55p; 8 × 6in 68p; 10 × 7in 80p; 12 × 8in £1; 4 × 9in £1; 20; 14 × 8in 60p; 16 × 10 in £1 40. ALUMINIUM PANELS 18 s.w.g. 6 × 4in 12p; 8 × 6 in 19p; 14 × 3in 20p; 19 × 7in 24p; 12 × 5in 52p; 12 × 8in 34p; 16 × 6in 34p; 14 × 9in 40p; 12 × 12in 47p; 16 × 10in 60p. 1] inch DIAMETER WAVECHANGE SWITCHES, 45p ea. 2 p. 2-way, or 2 p. 6-way, or 3 p. 4-way. TOGGLE SWITCHES, 4p. 20p; dp. 25p; dp. dt. 30p.

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4/850V	14p	500/25V	20p	32 + 32/350V	35p
8/350V	22p	100 + 100/27	5V 65p	32 + 32/450V	
16/350V	30p	150 + 200/27			
32/500V		8 + 8/450V		16 + 16 + 16/2	
		8+16/450V			
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100/25V	10p	32 + 32/350V	40p	4700/63V	95p
LOW VOL	TAGE	ELECTROL	YTICS.		
1, 2, 4, 5, 1	8, 16,	25, 30, 50, 10	0. 200m	F 15V 10p.	
500mF 19	V 15	: 25V 20p: 50	V 30p.		
1000mF 12	V 201	: 25V 35p; 50	V 47p;	100V 70p.	
2000mF 61	7 25p:	25V 42p; 50V	/ 57p.		
		: 3000mF 25			
5000mF 61	/ 25p;	12V 42p; 25V	75p: 8	35V 85p; 50V 9	95p.
				Mica 2 to 500	
PAPER 3	60V-0-	1 7p; 0.5 18	p: lmF	15p; 2mF 15	OV 15p.
500V-0-001	to 0.	05 4p; 0-1 10p	: 0.25	8p: 0-47 25p.	
		0-0" 208pF + 1			
				25pF + 25pF.	50p;
				in 120p <b>F 50</b> p.	
SHORT W	AVE	SINGLE. 25p	F, 45p	; 50pF, 55p.	
NEON PAR	(EL I	NDICATORS 2	50V A	C/DC. Amber ?	30p.
RESISTOR	S W	. 1W. 1W. 20	°, 2p:	2W. 10p. 10Ω	to 10M.
HIGH STA	BILIT	Y. 1W 20. 1	0 ohm	to 6 meg., 10	D.
		rred values 10			
				t. 10 watt. 1	5 watt.
10 ohms to				"	

TAPE OSCILLATOR COIL Valve type 35p. FERRITE ROD 8 × fin 20p; 6 × fin 20p; 3 × fin 10p

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Long spindles. Midget Size 5 K. ohms to 2 Meg. LOG or LIN. L/S 20p. D.P. 35p. STEREO L/S 55p. D.P. 75p. Edge 5K. S.P. Transistor 25p.

80 Ohm Coax 5p yd. BRITISH AERIALITE AERAXIAL-AIR SPACED 40 yd £2.00; 60 yd £3.00. FRINGE LOW LOSS O per Ideal 625 and colour. O pyd

Wire Wound controls 1; in diam, 3 Watts. 10 ohms to 100K British Made with long spindles it in dia. 85p each.
DUAL CONCENTRIC POT 500K LOG AND 500K LIN D.P.,
switch. Inner spindle 3in; outer spindle 2in 75p.

# E.M.I. $13\frac{1}{2} \times 8$ in. SPEAKER SALE!

With tweeter and crossover. 10 \$5.25 watt. State 3 or 8 or 15 ohm. As illustrated. Post 35p

With flared tweeter cone and ceramic magnet. 10 watt.

Bass res. 48-80 c/s.

Flux 10,000 gauss.

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18 × 8in Bass unit 20 watt rubber cone surround \$6-80

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GOODMANS 64in. HI-FI SPEAKER

8 ohm. 10W. Large ceramic magnet. Special Rubber cone surround. Frequency response. Twin cone. 30-12,000 c/s. Ideal P.A. 30-12,000 c/s. Ideal P.A. Columns, Hi-Fi Enclosure Systems, etc.





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The moving coil disphragm gives a good radiation pattern to the higher frequencies ranation pattern to the higher requestions and a smooth extension of total response from 1,000 c/s to 18,000 c/s. Size 3½ × 2in deep. Rating 10 W. 3 ohm. £2.20 Post 30p. Crossover £1-85

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8 ohm 15 watt. Rubber sur-round. Heavy ceramic magnet. Bass resonance 35 cps. Fre-quency response 30-8.000 cps. Ideal bass unit for Hi-Fi system.



# SPECIAL OFFER LOUDSPEAKERS

3 ohm, 2in; 2in; 3in; 5in; 7 × 4in; 8 × 5in. 8 ohm, 2in; 2in; 5in × 3in; 3in; 4in; 6ia; 6 × 4in. 15 ohm, 3in; 5in; 6 × 4in; 5 × 3in; 7 × 4in; 8 × 5in. 25 ohm, 2in; 5 × 3in; 5in; 6 × 4in; 7 × 4in. 35 ohm, 3in; 5in, 64 ohm 2in. 80 ohm, 2in; 2in, 120 ohm 3in.

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CASSETTE MACHINE MOTOR. 6 Volt. Will replace many types £1.25.

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Complete with 12 ft. twin lead fitted with din speaker plus. Ready assembled with leads for speakers, bass, mid and tweeter. Crossover frequencies—980 cps and £2-25 3.000 cps.

£20-00

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1.5V d.c. operation over 200 hours continuous on SP2 battery, fully adjustable swing and speed. Ideal displays, teaching electro magnetism or for metronome. 95p Post strobe, etc.

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2 stage triode pentode valve. 3 watts output. Volume on/off and tone controls. Printed circuit A.C. mains complete and tested. 44.50 35p Complete with speaker.

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£6.75 THE PAIR, Post 45p.
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Wooter £4.25; Tweeter

Comprising a fine example of a Wooler 10? × 6!in with a massive Geramic Magnet, 440c Gauss 13,000 lines. Aluminium Cone centre to improve middle and top response. Also the E.M.I. Tweeter 3 in square has a special

Tweeter 3in square has a special lightweight paper cone and magnet flux 10.000 lines. Crossover condenser and ull instructions supplied.

Impedance Standard 8 ohms Maximum power 12 watts
Useful Response 35 to 18,000 cps
Bass Resonance 45 cps
SUITABLE PRECIOUITE 90 12 2 18 Base Resonance 45 cps SUITABLE ENCLOSURE 20 × 13 × 12in, MODERN DESIGN, TEAK WOOD FINISH.



£12.50 Post 75:



# BAKER MAJOR 12" £11.50

30-14,500 c/s, 12in. double cone, woofer and tweeter cone together with a BAKER ceramic magnet assembly having a flux density of 14,000 gauss and a total flux of 145,000 Maxwells. Bass resonance 40 c/s Rated 20 watts. NOTE: 3 or 8 or 15 obns must he stated. 15 ohms must be stated.

Module kit, 30-17,000 c/s with tweeter, crossover, baffie and instructions. £14.50

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Robustly constructed to stand up to long periods of electronic power. As used by leading groups.
Useful response 30-13,000 cps.
Bass Resonance 55 cps.

GROUP "25" 12in 25 watt 3, 8 or 15 ohms.

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15in. 50 watt 8 or 15 ohms.







# MAJOR 100 WATT **ALL PURPOSE TRANSISTOR**

All purpose transistorised.
Ideal for Groups, Disco and P.A.
4 inputs speech and music. 4 way
mixing. Output 8/15 ohm a.c. Mains.
Separate treble and bass controls.
Guaranteed. Details S.A.E.

DE-LUXE MODEL IN WOOD CABINET, BLACK, £69.



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3 CHANNEL SOUND TO LIGHT KIT 1,000 watts per channel. As featured in £12.50

# R.C.S. STEREO DECODER

British made. Ready aligned and tested. Complete £4.95 with instructions. Size 3in × 2in.

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	P50/2CC P50/1AC P50/3CC PCA1	40p 60p 40p 85p		RA2W OPT1 LFDT4 Twin gang	85p 65p 65p £1.20

£1.85 Post 30p

E.M.I. GRAM MOTOR 120V or 240V a.c. 2,400 rpm. 2-pole 70mA, Size 2! × 2! × 2!in.

£1.25 Post 25p **BAKER HI-FI SPEAKERS** HIGH QUALITY BRITISH MADE REGENT

12in. 15 watts

An inexpensive unit for the beginner in high fidelity and for general purposes. May be used to improve any Radio, Amplifier, Hi-Fi or Television receiver.

Bass Resonance 45cps

Flux Density 12,000 gauss Useful response 45-13,000cps 3 or 8 or 15 ohm models,

£9.50 Post

# DE-LUXE Mk II 12in, 15 watts

Especially designed to provide full range reproduction at an economical cost. Suitable for use with any high fidelity system. Built-in concentric

system. Built-in concentric tweeter cone.
Bass Resonance 30cps
Flux Density Useful response 25-16,000cps 8 or 15 ohms models.

£12.50 Post

# **SUPERB** 12in. 20 watts

A high quality londspeaker, its remarkable low cone resonance ensures created to be cone the copper drive and concentric tweeter cone resulting in full range reproduction with remarkable efficiency in the upper register.

Bass Resonance Fiux Density 16,500 gauss Useful response 20-17,000cps Sor 15 ohns models.

Useful response 20-8 or 15 ohms models.

£17 Post

# AUDITORIUM 12in. 25 watts

A Iuli range reproducer for high power, Electric Guitars, public address, multi-speaker systems, electric organs, Ideal for Hi-Fi and Disco-

Idea: for hi-Fi and Disco-theques.

Bass Resonance
Flux Density 15,000 gauss
Useful response 25-16,000cps
8 or 15 ohms models.

£14.50 Post

# **AUDITORIUM** 15in 35 watts

A high wattage loudspeaker of exceptional quality with a level response to above 8,000 cps. Ideal for Public Address. Discotheques, Electronic instruments and the home Hi-Fi.

Bass Resonance 35cps
Flux Density 15,000 gauss
S or 15 ohms models.

Hi-Fi Enclosure Manual containing plans, designs, crossover data and cubic tables. 68p.



# 8" or 10" x 6" ELAC HI-FI SPEAKER

Dual cone plasticised roll sur-round. Large ceramic magnet. 50-16,000 cps. Bass resonance 55 cps. 8 ohm impedance. £4.35

10in round £5

TEAK VENEER HI-FI SPEAKER CABINETS Fluted Wood Fronts

This famous unit now available, 10 watts, 8 ohm.

MODEL "A".  $20 \times 13 \times 12$  in For 12 in. dia, or £12.50 Post 10 in speaker.

MODEL "B". 16 × 10 × 7in For 13 × 8in. or £7-60 Post 8 in. speaker £7-60

Bin. speaker

MODEL "46": 30 × 20 × 18in.

Reflex cabinet will accept 1-12in.

bass unit, 1-5in. mid range, 1-3in.

tweeter. Teak finish. Grooved

front

£21.50 Carr.

LOUDSPEAKER CABINET WADDING 18in wide, 20p ft.

**GOODMANS CONE TWEETER** 34in. diam. 18,000 C.P.S. 25 WATTS 8 12 23-60

BARGAIN 4 CHANNEL TRANSISTOR MONO MIXER. Add musical highlights and sound effects to recordings. Will mix Microphone, records, tape and tuner with separate controls into single output. 9 volt battery £5.20

£5·20 operated. LD'ZU STEREO VERSION OF ABOVE 26-85.

BARGAIN 3 WATT AMPLIFIER. 4 Transistor
Push-Pull Ready built with volume, treble and
bass controls. 18 volt battery operated. Mains Supply 23 45.

THE "INSTANT" BULK TAPE ERASER & HEAD DEMAGNETISER. Suitable for cassettes, and all sizes of tape reels. A.C. mains 200/250V, Leaflet S.A.E. 40.25 Post £4.35 Post 20p



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WAFER HEATING ELEMENTS
OFFERING 1001 USES for every type of beating and
drying applications in the home, garage, greenhouse
lactory (available in manufacturing quantities)
size 10½ × 8½ × ½ in. Operating voltage 200/250V. a.c.
250 watts approx. Printed circuit element enclosed in
absetsos fitted with connecting wires. Completely flexible
providing safe Black heat. British-made for use in photocopiers and print drying equipment.
Ideal for home handymen and experimenters. Suitable
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Must be clamped between two sheets of metal or arbestos,
etc., to make efficient clothes dryers, towel rais—ideal for
airing cupboards. Ideal for anti-frost device for the garage
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Use in greenhouse for seed raising and plant protection.
Invaluable aid for bird houses, incubators, etc., etc. Can
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Practical Electronics September 1975

Buses 50, 68, 159. Roil Selhurst

Telephone 01-684-1665









































Teak veneered jin thick wood cabinet.

Size 184 in × 184 in × 84 in. Weight
231b. This cabinet features a wide
mesh Silver Grill covering a separate
compartment for mounting Tweeters
or Mid-Range Horn. The fully sealed
bass compartment is cut out for
64 im Wooler. 27:50. Carr. 85p. Rosewood Version 28:50.

Carr. 85p. Baffle could be cut for larger speaker.

WEYRAD COILS

E.M.I. TAPE MOTOR
4 pole, 240 v. 135 mA. Size 3½ × 2½ × 2½in.
1200 rpm. Spindle ¼in. diameter.











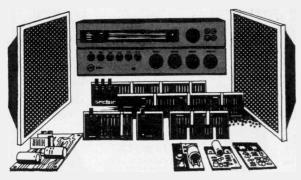








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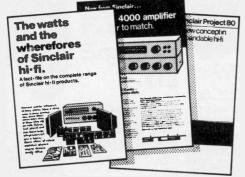
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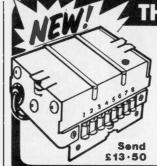
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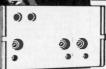
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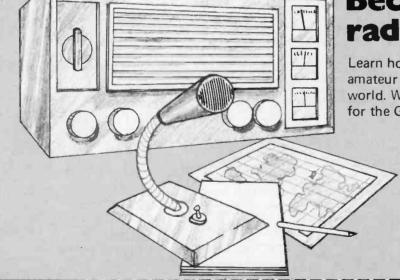
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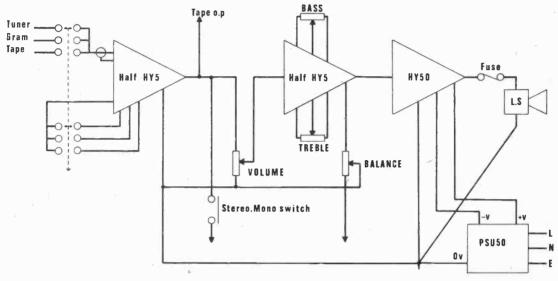
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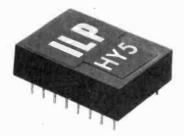
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AC125	0.18	ClilE	0.51	AF139	0.31	ST141	0.18	BF178	0.31	ı
AC126	0·18 0·19	C400 C407	0·31 0·26	AF178 AF179	0·51 0·51	TIP29 TIP30	0·44 0·52	BF179 BF180	0·31 0·31	ı
AC127 AC128	0.19	C424	0.26	AF180	0.51	TIP31A	0.56	BF181	0.31	ı
AC132	0.15	C425	0.51	AF181	0.51	TIP32A		BF182 BF183	0·41 0·41	L
AC134 AC137	0·15 0·15	C426 C428	0·36 0·20	AF186 AF239	0.51	TIP41A TIP42A	0.81	BF184	0.26	
AC141	0.19	C441	0.31	AL102	0.68	TIS43	0.31*	BF185	0.31	
AC141K	0.30	C442 C444	0.31	AL103 ASY26	0.88 0.26	UT46 ZN414	0.28*	BF187 BF188	0.28 0.41	ı
AC142 AC142K	0·19 0·26	C450	0.22	A8Y27	0.31	2G301	0.19	BF194	0.12	
AC151	0.16	MAT100	0.19	ASY28	0.26	2G302	0·19 0·19	BF195 BF196	0·12 0·15	
AC154 AC155	0.20	MAT101 MAT120	0·20 0·19	A8Y29 A8Y50	0.26	2G303 2G304	0.25	BF197	0.15	
AC156	0.20	MAT121	0.20	ASY51	0.26	2G306	0.41	BF200 BF222	0.46	L
AC157 AC165	0·25 0·20	MJE521 MJE2955	0.56	ASY52 ASY54	0·26 0·26	2G308 2N2926	0·36 G 0·13	BF257	0.98 0.48	
AC166	0.20	MJE3055	0.57	ASY55	0.26	2N2926	Y 0.11	BF258	0.61	
AC167 AC168	0.20	MJE3440 MPF102	0·51 0·43	ASY56 ASY57	0·26 0·26	2N2926 2N2926	O 0 10 R 0 10	BF259 BF262	0.87 0.56	L
AC169	0.15	MPF104	0.38	ASY58	0.26	2N2926	B 0.10	BF263	0.56	
AC176	0.20	MPF105	0.38	ASY73	0·26 0·41	2N3010	0·71 0·15	BF270 BF271	0·36 0·31	
AC177 AC178	0.25	OC19 OC20	0.36	ASZ21 BC107	0.08	2N3011 2N3053	0.18	BF272	0.81	
AC179	0.29	OC20 OC22	0.47	BC108	0.08	2N3054	0.47	BF273 BF274	0.36	
AC180 AC180K	0·20 0·30	OC23 OC24	0.49	BC109 BD136	0·08 0·41	2N3055 2N3319	0·42 0·15	BFW20	0·36 0·61	
AC181	0.20	OC25 OC26	0.39	B D137	0.46	2N3391	A 0.17	BFX29	0.28	
AC181 K AC187	0.30	OC26 OC28	0·30 0·51	BD138 BD139	0.51	2N3392 2N3393	0·15 0·15	BFX84 BFX85	0.22	
AC187K	0.22	OC29	0.51	BD140	0.61	2N3394	0.15	BFX86	0.22	
AC188	0.22	OC35	0.43	BD155	0.81	2N3395 2N3402	0·18 0·21	BFX87 BFX88	0·25 0·22	
AC188K ACY17	0.23	OC36 OC41	0·51 0·20	BD175 BD176	0·61 0·61	2N3402	0.21	BFY50	0.20	1
ACY18	0.20	OC42	0.25	BD177	0.67	2N3404	0.29	BFY51 BFY52	0.20	t
ACY19 ACY20	0·20 0·20	2N918 2N929	0·31 0·21	BD178 BD179	0.67 0.71	2N3405 2N3414	0·43 0·16	2G309	0.20	
ACY21	0.20	2N930	0.21	BD180	0.71	2N3415	0.16	2G339	0.20	
ACY22 ACY27	0.17	2N1131 2N1132	0.20	BD185 BD186	0.67 0.67	2N3416 2N3417	0.29	2G339A 2G344	0·17 0·19	١,
ACY28	0.19	2N1302	0.15	BD187	0.71	2N3525	0.77*	2G345	0.17	1
ACY29 ACY30	0.36	2N1303 2N1304	0.15	BD188 BD189	0·71 0·77	2N3614 2N3615	0.69	2G371 2G371B	0·17 0·12	1
ACY31	0.29	2N1305	0.18	BD190	0:77	2N3616	0.76	2G373	0.18	1
ACY34 BC173	0.21	2N1306	0·21 0·21	BD195 BD196	0·87 0·87	2N3646 2N3702	0·09 0·12	2G374 2G377	0·18 0·31	1
BC173	0·15 0·15	2N1307 2N1308	0.24	BD197	0.92	2N3703	0.12	2G378	0.17	1
BC175	0.22	2N1309	0.24	BD198	0.92	2N3704	0.13	2G381	0·17 0·17	
BC177 BC178	0·19 0·19	2N1613 2N1711	0.20	BD199 BD200	0.98	2N3705 2N3706	0·12 0·12	2G382 2G401	0 31	
BC179	0.18	2N1889	0.32	BD205	0.81	2N3707	0.13	2G414	0 31	
BC180 BC181	0.25	2N1890 2N1893	0·46 0·38	BD206 BD207	0·81 0·98	2N3708 2N3709	0.08	2G417 2N388	0.26	
BC182	0.15	2N2147	0.73	BD208	0.98	2N3710	0.08	2N388A	0.56	
BC182L BC183	0.15	2N2148 2N2192	0·58 0·36	BDY20 BF115	£1.02 0.25	2N3711 2N3819	0.09	2N404 2N404A	0·20 0·29	1
BC183L	0·15 0·15	2N2193	0.36	BF117	0.48	2N3820	0.51	2N524	0.43	
BC184	0.20	2N2194 2N2217	0.36	BF118 BF119	0·71 0·71	2N3821 2N3823	0·36 0·29	2N527 2N598	0·50 0·43	-
BC184L BC186	0·20 0·29	2N2218	0.20	BF121	0.46	2N3903	0.29	2N599	0.46	
BC187	0.29	2N2219	0.20	BF123	0.51	2N3904	0.31	2N696	0·18 0·14	١.
BC207 BC208	0·11 0·11	2N2220 2N2221	0.22	BF125 BF127	0·46 0·51	2N3905 2N3906	0·29 0·28	2N697 2N698	0.25	7
BC209	0.12	2N2222	0.20	BF152	0.56	2N 4058	0.12	2N699	0.36	7
BC212L BC213L	0·13 0·13	2N2368 2N2369	0·18 0·15	BF153 BF154	0·46 0·46	2N4059 2N4060	0·10 0·12	2N706 2N706A	0.08	1
BC214L	0.17	2N2369A	0.15	BF155	0.71	BC113	0.10	2N708	0.12	î
BC225	0·26 0·36	2N2411 2N2412	0·25 0·25	BF156 BF157	0·49 0·56	BC114 BC115	0·16 0·18	2N711 2N717	0·31 0·36	-
BC226 BC301	0.28	2N2646	0.48	BF158	0.56	BC116	0.16	2N718	0.25	
BC302 BC303	0·25 0·31	2N2711 2N2712	0·21 0·21	BF159 BF160	0·61 0·41	BC117 BC118	0·19 0·10	2N718A 2N726	0·51 0·29	
BC304	0.37	2N2712	0.21	BF162	0.41	BC118	0.31	2N727	0.29	
BC440	0.31	2N2904	0·18 0·21	BF163 BF164	0·41 0·41	BC120	0.81	2N743 2N744	0-20 0-20	1
BCY30	0·37 0·25	2N2904A 2N2905	0.21	BF165	0.41	BC125 BC126	0·12 0·19	2N914	0.15	
BCY31 BCY32	0.27	2N2905A	0.21	OC44	0.16	BC132	0.12	2N4061	0.12	3
BCY32 BCY33	0·31 0·22	2N2906 2N2906A	0·16 0·19	OC45 OC70	0·18 0·10	BC134 BC135	0·19 0·12	2N 4062 2N 4284	0·12 0·18	1
BCY34	0.26	2N2907	0.20	OC71	0.10	BC136	0.16	2N4285	0.18	1
BCY70 BCY71	0·15 0·20	2N2907A 2N2923	0·22 0·15	OC72 OC74	0·15 0·15	BC137 BC139	0·16 0·41	2N4286 2N4287	0·18 0·18	1
BCY72	0.15	2N2924	0.15	OC75	0.16	BC140	0.81	2N4288	0.18	1
BCY72 BCZ10	0.20	2N2925	0.15	OC76	0.16	BC141	0.31	2N4289 2N4290	0·18 0·18	13
BCZ11 BCZ12	0·26 0·26	ACY35 ACY36	0·21 0·29	OC77 OC81	0·26 0·16	BC142 BC143	0·31 0·31	2N4291	0.18	
BD115	0.63	ACY40	0.18	OC81D	0.16	BC145	0.46	2N 4292	0.18	1
BD116 BD121	0·81 0·61	ACY41 ACY44	0.19	OC82 OC82D	0·16 0·16	BC147 BC148	0·10 0·10	2N4293 2N5172	0·18 0·12	Г
BD123	0 67	AD130	0.39	OC83	0.20	BC148 BC149	0.12	2N5194	0.56	1
BD124 BD131	0·70 0·51	AD140 AD142	0·49 0·49	OC139 OC140	0.20	BC150 BC151	0·19 0·20	2N5294 2N5296	0·56 0·56	1
BD132	0.61	AD143	0.39	OC169	0-26	BC152	0.18	2N5457	0.32	
BD133 BD135	0.67	AD149 AD161	0·51 0·36	OC170 OC171	0·26 0·26	BC153 BC154	0·29 0·31	2N5458 2N5459	0·32 0·41	1.
BFY53	0.18	AD161	0.36	OC200	0.26	BC157	0.19	2N6122	0.69	1
BSX 19	0.16	AD161 &		OC201	0.29	BC158	0·12 0·12	28301 28302A	0·51 0·43	1
BS X 20 BS Y 25	0·16 0·16	AD162(M	P) 0-69	OC202 OC203	0.26	BC159 BC160	0.12	28302A 28302	0.43	1
BS Y26	0.16	. ADT140	0.51	OC204	0.26	BC161	0.51	28303	0.56	1
BSY27 BSY28	0·16 0·16	AF114	0.25	OC205 OC309	0.36 0.41	BC167 BC168	0·12 0·12	28304 28305	0·71 0·80	1
BS Y29 BS Y38	0.16	AF115 AF116	0·25 0·25	OCP71 ORP12	0.44*	BC169	0·12 0·12	28306 28307	0.80	1
BSY39	0.19	AF117	0.25	ORP12	0.41*	BC170 BC171	0.12	28321	0.75	(
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# **SEMICONDUCTORS**

\* 74 SERIES T.T.L. I.C's

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$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	BI-PAK	STIL	L LOW	VEST 1	IN PRIC	E. FUL	L SPI	CIFIC	ATION G	UARAN	TEED	. ALL
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	m.,							ORERE		_		
7400         0.14         0.13         0.12         7448         £1.02         0.99         0.97         74122         0.65         0.63         0.80           7401         0.14         0.13         0.12         7450         0.14         0.13         0.12         7462         0.14         0.13         0.12         7463         0.14         0.13         0.12         7461         0.14         0.13         0.12         7464         0.14         0.13         0.12         7464         0.14         0.13         0.12         7464         0.14         0.13         0.12         7464         0.14         0.13         0.12         7463         0.14         0.13         0.12         7460         0.14         0.13         0.12         7460         0.14         0.13         0.12         7460         0.14         0.13         0.12         74150         2.14         0.13         0.12         74150         2.14         0.13         0.12         74150         2.14         0.20         0.27         0.25         74153         2.09         0.93         0.88         0.83           7407         0.36         0.31         0.29         7470         0.30         0.27         0.25 <th< td=""><td>Type</td><td></td><td></td><td></td><td>Type</td><td></td><td></td><td>100</td><td>Type</td><td>Qua</td><td></td><td>100</td></th<>	Type				Type			100	Type	Qua		100
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	7400				m. 1 in					1		
7402         0-14         0-13         0-12         7451         0-14         0-13         0-12         7461         0-14         0-13         0-12         7461         0-14         0-13         0-12         74145         9-12         9-16         6-17           7404         0-14         0-13         0-12         7464         0-14         0-13         0-12         9-16         21-18         21-20         24-20           7405         0-14         0-13         0-12         7460         0-14         0-13         0-12         7451         21-20         0-92         0-97         7407         0-30         0-27         0-25         74153         0-98         0-88         0-83           7407         0-30         0-27         0-25         74154         9-16         9-17         0-88         0-83         0-88         0-83         0-88         0-83         0-80         0-88         0-83         0-80         0-82         74154         9-16         21-14         10-16         0-16         0-16         0-17         0-17         0-86         0-84         0-52         74156         21-11         21-06         21-02         74161         0-18         0-18         0-18												
7403         014         013         012         7463         014         013         012         7463         014         013         012         74150         £1.20         £1.11           7404         014         013         012         7460         014         013         012         74150         £1.39         £1.30         £1.20         £9.97         0.93         0.82         7480         0.94         0.94         0.94         0.13         0.12         74615         £1.39         £1.30         £1.20         0.97         0.93         0.82         0.83         74155         £1.11         £1.06         £1.02           7410         0.14         0.13         0.12         7475         0.56         0.64         0.52         74157         61.33         £1.11         £1.06         £1.02         74157         61.03         £1.11         £1.06         £1.02         74157         61.03         £1.41         £1.02         £1.02												
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7442         0.69         0.68         0.59         74107         0.41         0.39         0.37         74194         £1.92         £1.16         £1.11           7443         £1.11         £1.08         £1.02         74110         0.58         0.51         0.46         74195         £1.02         0.97         0.93           7444         £1.11         £1.08         £1.02         74111         0.83         0.81         0.78         74196         £1.11         £1.08         £1.02           7446         £1.11         £1.08         £1.02         74118         0.83         0.88         0.83         74197         £1.11         £1.06         £1.02           7446         £1.11         £1.08         £1.02         74118         £1.38         £1.30         £1.20         74190         £2.55         £2.50												
7443 21.11 21.08 21.02 74110 0.58 0.51 0.48 74195 21.02 0.97 0.93 7444 21.11 21.08 21.02 74115 0.83 0.81 0.78 74196 21.11 21.08 21.02 7415 21.48 21.44 21.39 74118 0.93 0.88 0.83 74197 21.11 21.08 21.02 7416 21.11 21.08 21.02 7419 21.38 21.38 21.30 21.20 74190 22.55 22.50 22.45												
7444 21.11 21.06 21.02 74111 0.83 0.81 0.78 74196 21.11 21.06 21.02 7445 21.48 21.44 21.39 74118 0.93 0.88 0.83 74197 21.11 21.06 21.02 7446 21.11 21.06 21.02 74119 21.38 21.30 21.20 74190 22.55 22.50 22.45												
7445												
7446 21:11 21:06 21:02 74:19 21:39 21:30 21:20 74:190 22:55 22:50 22:45												
7447 £1.02 0.99 0.97   74121 0.46 0.44 0.41   74199 £2.31 £2.21 £2.11												
	7447	£1.02	0.99	0.97	74121	0.46	0.44	0.41	74199	£2·31	£2.21	£2.11

Devices may be mixed to qualify for quantity price. (TTL 74 series only) data is available for the above series of I.C's in booklet form. PRICE 35p.

*				D.T.L	930	SE	RIE	S			
Type			intities	Type	Qua	ntities		Type	Quan		100 1
	1	25	100 +		1	25	100 +		1	25	100+
BP930	0.14	0.18	0.12	BP944	0.15	0.14	0.13	BP962	0.14	0.13	0.12
BP932	0.15	0.14	0.13	BP945	0.28	0.26	0.23	BP9093	0.42	0.40	0.38
BP933	0.15	0.14	0.13	BP946	0.14	0.13	0.12	BP9094	0:42	0.40	0.38
BP935	0.15	0.14	0.13	BP948	0.28	0.26	0.23	BP9097	0.42	0.40	0.38
BP936	0.15	0.14	0.13	BP951	0.65	0.60	0.56	BP9099	0.42	0.40	0.38

T8014 14 pin type 0.31 0.28 0.25
TS016 16 pin type 0.35 0.32 0.30
TS020 24 pin type 0.35 0.32 0.30
TS020 24 pin type (0.69 0.64 0.62
BPS 8 8 pin type (low cost) 0.14 0.12 0.10
BPS14 14 pin type (low cost) 0.15 0.13 0.11
BPS16 16 pin type (low cost) 0.16 0.14 0.12

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Type	Qua	ntities		Type	Qui	antities	- 3	Type	Qu	antities	
	1	25	100+		1	25	100+		1	25	100 +
72702	0.46	0.44	0.42	SL701C	0.46	0.42	0.37	uA723C	0.45	0.43	0.40
72709	0.23	0.21	0.19	SL702C	0.46	0.42	0.37	76003	£1·39	£1·34	£1.30
72709P	0.19	0.18	0.17	TAA263	0.74	0.65	0.56	76023	£1·39	£1·34	£1.30
72710	0.82	0.31	0.28	TAA293	0.93	0.88	0.83	76660	0.88	0.86	0.83
72741	0.28	0.27	0.26	TAA350	A			LM380	0.93	0.90	0.88
72741C	0.26	0.25	0.24		£1.71	£1.67	£1.57	N E555	0.45	0.43	0.40
72741P	0.28	0.27	0.26	uA703C	0.26	0.24	0.22	NE556	0.88	0.86	0.83
72747	0.79	0.74	0.61	uA709C	0.19	0.18	0.17	TBA800	£1.39	£1.34	£1.30
72748P	0.35	0.33	0.31	uA711C	0.32	0.31	0.28	ZN414	£1-11	-	
8L201C	0.46	0.42	0.37	uA712C	0.32	0.31	0.28				

SILICON	RECTIFIERS

	300mA	750mA	1 Amp	1.5	5Amp	3Amp	10Amp	30 Amp
PIV	(DO 7)	(80 16)	Plastic	(80	16)	$(80\ 10)$	(80 10)	(TO 48)
50	0.05	0.06	IN 4001	0.05	0.07	0.14	0.19*	0.56*
100	0.05	0.07	1N4002	0.06	0.09	0.16	0.21*	0.69*
200	0.06	0.09	1N4003	0.07	0.12	0.20	0.23	0-93*
400	0.07	0.14	1N4004	0.08	0.14	0.28	0.35*	£1.25*
600	0.08	0.16	1N4005	0.09	0.16	0.33	0.42*	£1.76*
800	0.11	0.18	1N4006	0.10	0.18	0.35	0.51*	£1·94*
1000	0.13	0.28	1N 4007	0.11	0.23	0.44	0.60*	£2.31*
1000		0.30			0.28	0.54	0.69*	£2-88*

# PO BOX 6 WARE HERTS

# SUPER UNTESTED PAKS

Pak :	W 0	Description	_	
			,	Pri
U 1		Glass Sub-min, General purpose Germ, dlodes	٠	0-0
U 2		Mixed Germanium translators AF/RF		0-
1,3	75	Germanium gold bonded sub-min, like OA5, OA47		0-0
U 4		Germanium transistors like OC81, AC128		0.0
17.5		200mA sub-min. silicon diodes		0.0
U 6	30	Sil. Planar trans. NPN like BSY95A, 2N706		0.6
U 7	16	8il, rect. TOP-HAT 750miA VLTG, RANGE up to 100		0-6
l. 8		Sil. planar dlodes DO-7 glass 250n <sub>1</sub> A like OA200/202	٠.	0-6
11 9		Mixed voltages, 1 Watt Zener Diodes		0-6
1710		BAY50 charge storage diodes DO-7 glass	٠.	0.6
UII		PNP 8ll. planar trans. TO-5 like 2N1132, 2N2904	٠.	0-6
₹113		PNP-NPN Sil. transisters OC200 & 28104		0.6
U14		Mixed silicon and germanium diodes	٠.	0.6
U15	20	NPN Sil. planar trans. TO-5 like 2N696, 2N697		0.6
U16		3Amp sll. rectifiers stud type up to 1000 PIV		0.6
1717		Germanium PNP AF transistors TO-5 like ACY 17-22		0.6
U18		6 Amp sli. rectifiers BYZ13 type up to 600 PIV		0.6
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20	0.15	0.18		-			***	_	_	
30	0.19	0.22	_	-		_	-	_		
50	0.22	0.28	0 20	0.25	0.36	0.36	0.48	0.51	0.54	£1·18
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AAY30	0.09	BY124	0-12	BYZ19	0.28	OA95	0.07
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BA116	0.21	BY128	0.16	CG651 (C		8D10	0.08
BA126	0.22	BY130	0.17	OA79)	0.07	8D19	0.06
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BA 155	0.15	BYX38/	30 <b>0·48</b>	OA10	0.14	1N914	0.08
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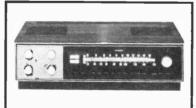
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# PRIVATE INNOVATORS

WHAT other branch of applied science procreates ideas as abundantly as electronics? Considering the number as well as the variety of interesting and useful applications it has fostered, electronics must stand supreme. This fertility has made electronics a household word. Familiar in a superficial sense through its widespread involvement with ordinary everyday affairs, electronics is still mysterious though to the greater part of the populace so far as its innermost workings are concerned.

Because of this technical barrier, it probably remains the popular view that all worthwhile ideas and inventions in the realm of electronics must originate within industry or other professional establishments. This is obviously true of most of the major and more significant developments in the history of applied electronics. Yet the great achievements of organised research and development, whether in private enterprise or public service areas, must not be allowed to overshadow or obliterate the genuine services rendered by private inventors in furthering the wider and often more down-to-earth use of electronic devices, circuits and systems.

One of the strengths of electronics arises from the fact that the subject is no closed shop. Throughout its history electronics has always attracted private experimenters and innovators. The amateurs of the earlier days were comparatively few and many of them showed remarkable devotion as well as rare technical astuteness in pursuing a new and little understood subject, and one that was then hemmed in within narrow confines. With subsequent development and progress, the field opened up increasingly for private endeavour. The coming of solid state devices finally produced the great breakthrough and enabled electronics to become established as the widely based technology of limitless utility it is today.

With this broadening out of the functions of applied electronics the opportunities and scope for the private individual likewise increased enormously. Now that electronics had become a tool of infinite application, the amateur could enter into the innovation area as never before; his awareness of common, or not so common, needs and problems in ordinary life leading to the harnessing of circuits to previously unheard of or undreamt of uses. And the past couple of decades have certainly shown how amateur imagination can play a valuable part in the full cultivation of what is potentially the most useful of all practical sciences.

It is therefore highly gratifying whenever industry shows that it recognises the value of the vast "Think Tank" that amateurs collectively can provide. Practical Electronics is especially delighted that one of Britain's top electronics firms, The Plessey Co. Ltd., has now joined with this magazine in sponsoring a unique kind of competition intended to tap this "Think Tank". Our readers are not likely to be short of ideas, and we look forward to a bumper crop of novel suggestions for useful and viable applications of a given circuit. Substantial rewards are waiting to be won, which should help to stir the imagination.

F.E.B.

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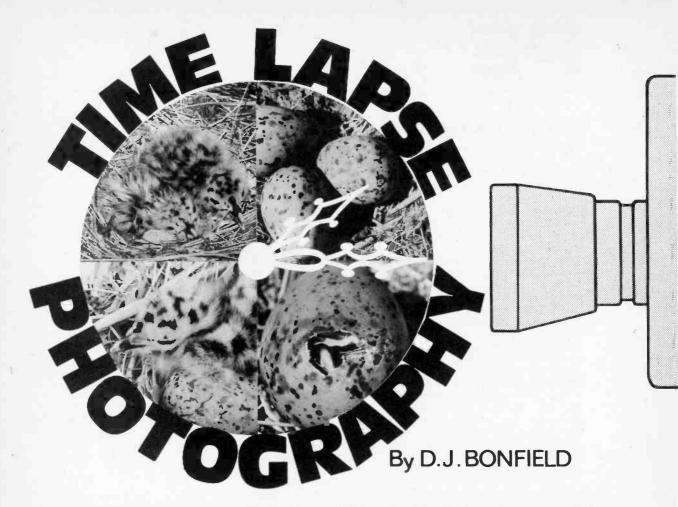
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For many years, time lapse photography has been an invaluable aid to the scientist, but to the ordinary photographer armed with a cine camera, it can open up a whole new field of beautiful and often strange effects. For example, a green rose bud may be transformed into a magnificent deep red rose in a matter of seconds, while at the other end of the scale, a forest of mustard and cress may appear as if by magic to rise from a seething mass of seeds.

But time lapse photography is by no means confined to plant life. Sunrises and sunsets provide good subjects, and even mould growing on a slice of damp

bread can be quite fascinating!

### **BASIC PRINCIPLES**

Any cine camera with the facility for exposing a single frame of the film with each operation of the shutter release. can be used in conjunction with the unit to be described, to effectively compress events lasting hours or even days into a few seconds on film.

The usual way to produce this effect is to reduce the speed of the film while in the camera, and project it back at normal speed. Consider the common 8mm cine camera, which usually runs at a speed of about 16 frames per second. If we wanted to compress 24 hours of filming into 15 seconds, we would want to take a

total of  $16 \times 15 = 240$  frames in 24 hours, which is one frame every 6 minutes.

To meet such requirements, the Time Lapse Controller was designed to provide a reliable automatic method of operating the camera at pre-set intervals. Many subjects to be photographed will require artificial lighting, and since leaving photoflood lights on continuously is wasteful, the facility has been incorporated in the unit to electronically switch up to 1kW of lighting throughout the period of each exposure.

The unit is connected to the camera via a cable release, and the time between successive exposures can be varied continuously over the range 10 seconds to 48 minutes.

# CIRCUIT DESCRIPTION

The complete circuit of the Time Lapse Controller is shown in Fig. 1. The requirements for this circuit are that during each cycle it should first switch on the lighting, and then allow approximately three seconds for the lights to reach their full intensity, after which a solenoid is to be operated for about a second to release the camera shutter, followed a second or so later by the lights switching off.

### 555 TIMER

The basic timing element of the circuit is the NE555V integrated circuit, IC1, which is used in an astable mode. The mark-to-space ratio of the

squarewave output can be varied considerably by the two controls S1 and VR1; one cycle is shown in Fig. 2.

When the output of IC1 is low, TR2 is driven to saturation, switching on the indicator l.e.d. D5 and the triac CSR1. This in turn switches on the photoflood lighting. (A triac conducts with either a positive or negative gate signal regardless of the voltage polarity across its MT1 and MT2 terminals.) The indicator l.e.d. is useful as it shows when the solenoid is about to operate when no lighting is plugged into the unit. Thus the lights are on for about 5 seconds, determined by the pre-set potentiometer VR3, during each cycle.

# DELAY

TR3 is an inverting buffer stage. As soon as the collector voltage rises, C3 begins to charge up through R22 and VR4. Due to the voltage drop across the silicon diode D6, TR4 does not begin to conduct until the voltage across C3 exceeds about 1 volt.

The Schmitt trigger input to the monostable IC2 is used so that the voltage at TR4 emitter has to reach a certain threshold voltage before the monostable is triggered. Thus the purpose of C3, D6, TR4, and associated resistors, is to delay by about 3 seconds the triggering of the monostable, once the lighting has been switched on.

Once triggered, the monostable remains in its unstable state for about a second, determined by C4 and VR5. During this period TR5 is saturated, turning on the triac CSR2, which operates the mains solenoid.

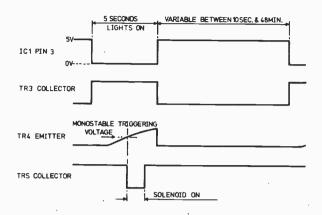


Fig. 2. Waveforms for one cycle of operation

The solenoid actuates the camera shutter via a cable release. Shortly afterwards the output of IC1 goes high, switching off the lights and completing the cycle.

# **POWER SUPPLY**

TR1 and associated components, operating in an emitter-follower arrangement, form a simple stabilised power supply, giving approximately 5 volts from a 9-0-9 volt 100mA miniature transformer. The 1.e.d. D4 is the mains-on indicator.

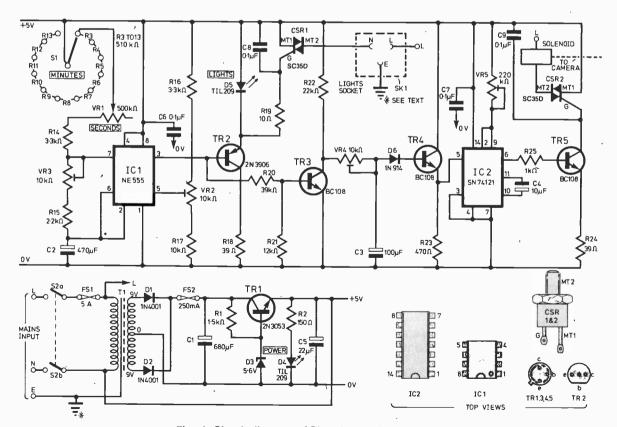
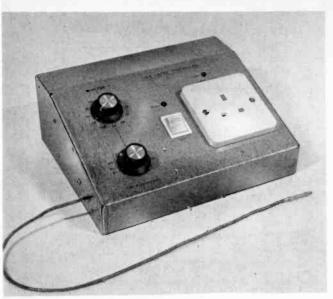


Fig. 1. Circuit diagram of Time Lapse Controller



Front panel arrangement of the time lapse controller

The circuit takes its maximum current of about 180mA when both triacs are conducting. With both triacs off, the quiescent current is in the region of 50mA.

# COMPONENTS

The triac CSR1 controlling the lights is rated at 6A, r.m.s., at ±400V at 75 degree C, so it is capable of switching about 1.4 kilowatt of lighting at mains voltage. It would probably be wise to consider one kilowatt the maximum, since no heat sink is used. When switching on one kilowatt for 5 seconds in every 10, the triac does not get sufficiently hot to warrant the use of a heat sink.

It should perhaps be emphasised that good quality components should be used for the timing capacitors. in particular C2. The higher the leakage current of this capacitor, the less accurate will be the timing on the higher ranges of S1, rather defeating the object of using 2 per cent tolerance resistors.

# CONSTRUCTION

The majority of the components are mounted on one printed circuit board measuring  $7\text{in} \times 3\text{in}$  (178 × 76mm), shown full size in Fig. 3, and the component layout is shown in Fig. 4.

The author advises the use of sockets for the i.c.s if only to provide a simple method of testing other 555's

and 74121's

The usual care must of course be taken with the orientation of the semiconductors and capacitors. A TO5 cooling clip is essential for TR1.

# TRIAC MOUNTING

The triacs are mounted through ‡in holes drilled in the circuit board. The cases of the triacs will be live when the unit is in operation, so it is essential to ensure that there is adequate space between adjacent components, tracks on the board, and the case.

Fairly thick wire is used to join the main terminal 1 (MT1) of CSR1 to MT1 of CSR2, and a further short length of wire to join these MT1's to the + 5 volt track

on the circuit board.

# COMPONENTS . . .

### Resistors 1.5kΩ R<sub>1</sub> $150\Omega$ R2 510kΩ2% R3-R13 $3.3k\Omega$ R14 $2.2k\Omega$ R15 3-3kΩ R16 $10k\Omega$ R17 39Ω ±W R18 R19 $10\Omega$ 30kO R20 R21 $12k\Omega$ $22k\Omega$ R22 R23 $470\Omega$ R24 $39\Omega$

All 1 watt 5% carbon unless otherwise stated.

# **Potentiometers**

1kΩ

VR1 -  $500k\Omega$  lin. VR2-4  $10k\Omega$  S/Min. Horiz. Preset VR5  $220k\Omega$  S/Min. Horiz. Preset

### Capacitors

R25

C1	$680 \mu F$	16V	elect.	C5 22µF 16V elect.
C2	470µF	6-3V	elect.	C6-9 0·1µF 16V
C3	100μF	10 V	elect.	polyester.
C4	10. F	10V	elect	

# Semiconductors

TR1 2N3053
TR2 2N3906
TR3-5 BC108
IC1 NE555V
IC2 SN74121
D1-2 1N4001
D3 5.6V 400mW Zener
D4-5 TIL209 l.e.d. + clip
D6 1N914
CSR1 SC35D 3A 400V triac (Henry's Radio)
CSR2 SC40D 6A 400V triac (Henry's Radio)

## **Switches**

S1 1-pole 12-way rotary
S2 Double-pole on/off rocker switch (5A min.)

### Solenoid

Westool type MM6, mains operated solenoid. (From Electro-Tech Components Ltd., 315/317, Edgware Rd., London W2)

### Case

Instrument case size 235mm  $\times$  190mm  $\times$  90mm type U. (From H. L. Smith & Co. Ltd., 287/289 Edgware Rd., London W2 1BE)

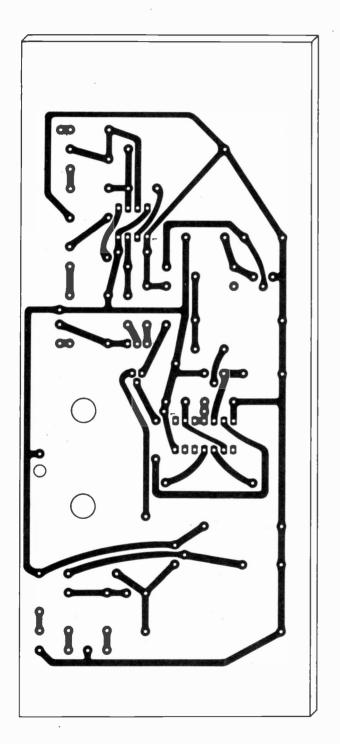
### Transformer

T1 Miniature mains transformer. 9-0-9V 100mA

### Miscellaneous

I.C. sockets; T05 cooling clip; two 20mm fuse holders; 5A fuse; 250mA fuse; 13A flush wall mounting socket; Bulgin mains connector (plug & socket); two knobs, cable release; p.c. board; aluminium for brackets; s.r.b.p. board and hardware for mounting solenoid.

# **CIRCUIT BOARD DETAILS**



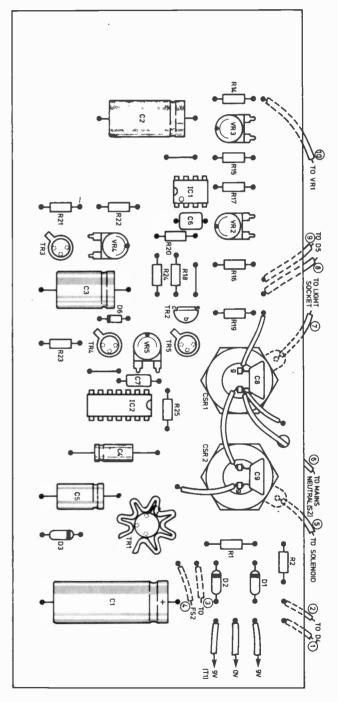
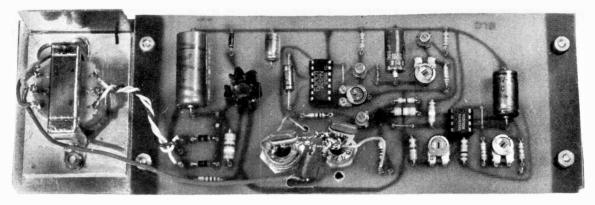


Fig. 3. Printed circuit board master

Fig. 4. Component layout



Components mounted on the printed circuit board and positioning of the transformer on the L-shaped heatsink

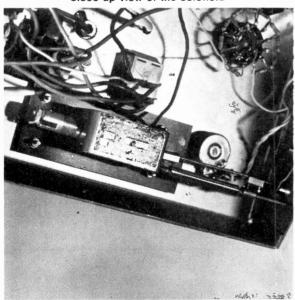
# CASE

The whole unit is housed in a  $9\frac{1}{4} \times 7\frac{1}{2} \times 3\frac{1}{2}$  in  $(235 \times 190 \times 89 \text{mm})$  aluminium case with a sloping front panel and an attractive hammered silver finish. Fig. 5 shows the layout inside the case. The front panel can be finished off with Letraset and protected by a coating of clear lacquer to give a professional appearance.

# INTERIOR LAYOUT

The circuit board is fixed vertically at the deep end of the case, with the components facing the ventilation slot. The C2 end of the board is attached to the side of the case by a  $\frac{1}{2} \times \frac{1}{2} \times 3$  in  $(12 \times 12 \times 76$ mm) long "L" shaped bracket made from a piece of aluminium sheeting. The opposite end is held by a  $\frac{1}{2} \times 2\frac{1}{2} \times 3$  in  $(12 \times 63 \times 76$ mm) long "L" shaped bracket. This second bracket provides a convenient position on which to mount the transformer. In the prototype a small plate of aluminium with a flange was sandwiched between the transformer and bracket to help dissipate any heat produced by the transformer.

Close-up view of the solenoid



# **EARTHING**

It must be emphasised that although the case is earthed, neither the positive nor the negative supply lines are connected to earth, but the positive supply line is connected to mains neutral.

# **COMPONENT POSITIONING**

The resistors R3 to R13 are mounted directly on the single-pole 12-way rotary switch S1. The two light emitting diodes are held in clips mounted in holes in the positions shown in Fig. 5. The mains on-off switch S2 is a double-pole rocker type rated at 5A or more. A Bulgin mains connector (chassis mounting plug and line socket) was used to give the unit a neater overall appearance. This can be mounted on one side of the cas together with the two 20mm fuse holders.

# WIRING

Insulated wire capable of carrying 5A must be used to connect the MT1's of the triacs to mains neutral via the switch S2. A standard 13A flush wall mounting socket is mounted on the front panel, into which the lighting can be plugged as required. 5A wire must again be used to connect MT2 of the triac CSR1 to the terminal marked neutral on this socket, and again from the terminal marked live on the socket to the mains live.

# THE SOLENOID ASSEMBLY

The solenoid recommended for this project has a pull of 3lb, which should be sufficient for even the most stubborn shutter release.

Fig. 6 shows the method used to connect the solenoid to the cable release. It is important to be able to adjust the probe depth setting so that the camera operates smoothly, and is not damaged by the probe depth being too great. This aim is not too difficult to achieve with the specified solenoid, making use of the 4BA tapped holes in the base and end of the solenoid housing.

# **PLATFORM**

First a platform is made out of a piece of s.r.b.p. or Perspex, approximately  $4\frac{1}{2} \times 1\frac{3}{4}$  in  $(115 \times 45 \text{mm})$ . Two slots about  $\frac{3}{4}$  in (19 mm) long have now to be made in it to correspond with the two tapped holes in the solenoid base. The slots are best made by drilling a series of  $\frac{1}{4}$  in holes and then filing down the jagged edges. The solenoid can then be attached to the

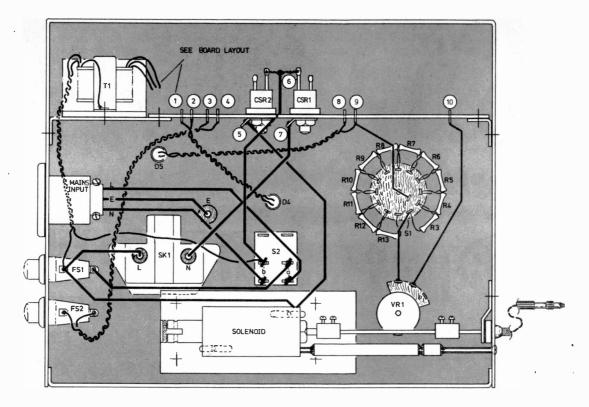


Fig. 5. Interior layout and interwiring details

platform by two short 4BA bolts. These bolts should not be screwed in all the way but be just loose enough to allow the solenoid to move reasonably smoothly up and down the slots.

Before the unit is put into operation, it will be necessary to glue the ends of the bolt shanks into the tapped holes with a strong adhesive such as Araldite, otherwise they may tend to shake loose due to the 50Hz vibration of the solenoid. The platform can now be mounted at each corner on a short pillar or a couple of nuts, in the position shown by Fig. 5.

The solenoid is moved along the platform slots by a short length of 4BA studding, one end of which is screwed through a tapped hole at the end of the solenoid. A 4BA bolt is pushed through a hole in the side of the case in line with the studding, to which it is attached by means of a 4BA tapped pillar or a spacer filled with plenty of Araldite. Thus by turning the head of the bolt, the solenoid can be moved towards or away from the side of the case.

# **CABLE RELEASE**

The next step is the cable release, which should be as long as possible; 20in should be considered the minimum. The button at the top of the cable release shaft has to be removed, with some cable releases it simply screws off, with others it will have to be sawn off.

A suitably sized hole is drilled in the side of the case in line with the solenoid plunger, so that the collar of the cable release fits snugly into it. The gap between the solenoid and cable release is best bridged with a thin metal rod. Some type of connectors are required to join the cable release to the rod to the solenoid.

Parts from a dismantled terminal block provide a good solution, but probably Araldite will again be required to stop them working loose.

# **SETTING UP**

Before switching on, the mains wiring must be carefully checked to ensure that the live and neutral have not become interchanged. A neutral wire should lead to the triacs, and a live wire to the solenoid and the lights socket.

If all is well, set all preset potentiometers to approximately their mid-positions. Set S1 to position 1, and set VR1 to minimum resistance. Replace FS1 by a 1A fuse for the initial testing, and plug a low wattage lamp such as a table lamp into the lights socket.

Check that the +5V or 0V supplies are not connected to earth, and ensure that the case is earthed.

Remembering that the cases of the triacs will be live, switch on. Within about 15 seconds, the table lamp should come on for several seconds together with the lights indicator, l.e.d. D5. Sometime during the period that the lamp is on, the solenoid should be actuated for about a second. This cycle should be repeated at approximately 10 second intervals.

# **PRESETS**

Preset VR3 can now be adjusted until the period during which the lights are on in each cycle lasts 5 seconds. Now adjust VR4 to produce a delay of

continued on page 719

# REE ENTRY Practical Electronics & Plessey issue a challenge IW INVENTIVE ARE YOU?

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PRIZES

★ 1st Prize £250

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★ 25 Magispark Gas Ignitors

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When you saw the Portable Gas Ignitor in the July issue of Practical Electronics did you immediately think of other applications for this particular circuit? Can you suggest any ways in which the device could be put to other uses in industry or in the home? If so, here's an opportunity to put your ideas to good advantage. And if you missed the July issue, not to worry—the circuit and the essentials of its operation are given on this page.

In association with the Plessey Company, we present this fascinating challenge and offer prizes totalling over £500 in value for really practicable and ingenious ideas. And a chance, possibly, to see your idea in production! So get that grey matter plus your know-now of electronics working, and meet our challenge right away!

# **HOW TO ENTER**

The contest is for practical applications of the Portable Gas Ignitor Circuit—either utilising the original circuitry or including design modifications to increase its scope.

To remind you, the operation of the Ignitor is:

The application of a high voltage across a pair of electrodes produces an electric field in the gas between them, this leads to ionisation and breakdown of the gas producing a spark across the gap. When the circuit of the ignitor is completed current flows from a battery into a transistor oscillator circuit and the resultant pulses of energy charge a capacitor to approximately 300 volts. The capacitor is then discharged by an electronic switch into the primary of an H.T. transformer which produces the necessary voltage to cause the sparks at the electrode. The design enables a steady stream of sparks at 10,000 V to be generated with complete safety and employing only a single 1.5V dry cell as the power source.

Entries must be written/drawn clearly on one side of plain paper with the entrant's full name and address at the top of every sheet. Each entry to comprise:

(a) a brief summary of the idea (about 25 words)

(b) any such further lucid description, drawings, sketches or circuit diagrams you consider the judges may need to form the best appraisal of your idea. DO NOT send actual models.

Each entry must have a properly completed entry coupon firmly affixed to the BACK of the summary.

### SECOND CHANCE!

Another free entry coupon will appear in the October issue of Practical Electronics.

The closing date is Monday October 13, 1975, to allow plenty of time for you to obtain the second entry coupon from our next issue and post two different ideas in one envelope if you wish.

### **RULES AND CONDITIONS**

There is no entry fee nor limit to the number of entries a reader may submit but each entry must be accompanied by a proper printed entry coupon. cut from PRACTICAL ELECTRONICS, and must bear the entrant's own full name and address. Entries will also be accepted from groups—in which case the entry coupon must be completed by one of the group and the names and addresses of all the other members listed on a separate piece of paper affixed, with the entry coupon, to the back of the summary.

All accepted entries will be examined by a panel of expert judges, including Plessey engineers and the Editor of *Practical Electronics*, and asssesed on (a) originality of the idea, (b) technical merit, (c) practicability, (d) economic viability, (e) market potential. The prizes will be awarded for the best entries in order of merit. No entrant may win more than one award. In the event of the same idea being submitted by two or more entrants, presentation of the entry (clarity, best expression, etc) will decide such winner(s) or winning order.

In the event that the judges consider there are not enough entries of a sufficiently high standard, the Editor reserves the right not to award any prize(s) at his discretion.

Entries arriving after closing date will not be considered, nor will any received that are illegible, not wholly understandable, are not accompanied by a properly completed entry coupon or in any other way do not comply exactly with the instructions and rules.

No responsibility can be accepted for entries lost or delayed in the post or otherwise; proof of posting will not be accepted as proof of receipt. No entries can be returned,

Copyright of all entries shall become the property of IPC Magazines Ltd., publishers of Practical Electronics. Ideas submitted may be used or adapted by the competition sponsors for production or other commercial use. Where appropriate, additional payment will be automatically negotiated with the entrant. Entries will not be published prior to evaluation in order to comply with legal safeguards.

Decisions of the judges, and of the Editor in all other matters affecting the competition, will be final and legally binding. No correspondence will be entered into nor interviews granted.

Winners will be notified by post and brief details of winning entries published later in *Practical Electronics*. The Editor reserves the right to amend and/or re-draw any sketches or diagrams of prizewinning entries for publication purposes.

The contest is open to all readers in Great Britain, Northern Ireland, Eire, Channel Isles and Isle of Man except employees of IPC Magazines Ltd., the Printers of Practical Electronics, and the Plessey Co. Ltd. and its subsidiary companies; and the families of all such employees.

Post your entry in a sealed envelope to: Ignitor Application Contest, PRACTICAL ELECTRONICS, 136 LONG AGRE, LONDON, WG2E 9QP, to arrive not later than October 13, 1975 the closing date.

Block Letters)
ADDRESS
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Felephone Number, if any
certify that
* delete clause NOT applicable
(a) this entry is of my own original idea and has not been copied from any other source.
(b) this entry is made on behalf of the group members listed, and is our own original idea not copied from any other source.
(c) this idea has not been published—or offered for publication elsewhere.
We agree to abide by the rules and conditions, and o accept the published result as final and legally inding.

# TIME LAPSE PHOTOGRAPHY

continued from page 717

3 seconds between the lights coming on and the solenoid starting to operate. Next adjust VR5 so that the solenoid operates for a period of one second, leaving one second between the solenoid switching off and the lighting switching off. Thus the camera shutter will be actuated only after the lighting has had time to reach its full intensity.

Now turn S1 to position 2, and with VR1 at minimum resistance, adjust VR2 until the time between subsequent cycles is exactly 4 minutes and 10 seconds. S1 will then be calibrated in 4 minute steps, and VR1 will be variable over the range 10 seconds to 4 minutes.

When the calibration has been completed, the 5A fuse FS1 can be replaced, and the unit tested with larger loads up to a maximum of one kilowatt.

# **CONNECTING UNIT TO CAMERA**

Turn the probe depth setting screw on the side of the case anticlockwise to push the solenoid to the far end of the slots, and connect the cable release to the camera. Now slowly screw up the probe depth setting screw, with the solenoid operating, until the setting is just sufficient to operate the camera shutter. Take care not to have the setting too deep or damage may be caused to the camera mechanism.

### THE UNIT IN OPERATION

If H hours of filming is to be compressed into s seconds of projection time, then the time between frames required is:

$$\frac{60 \times H}{P \times s}$$
 minutes

where P is the speed of the film when projected (i.e. 16 or 18 frames per second).

If the filming is to continue throughout the day and night, the light falling on the subject from the photoflood lighting must be of far greater intensity than any daylight that may reach the subject. Otherwise the exposure of the finished film will be inconsistent. Artificial light film should of course be used.

Now with a certain amount of patience (since it takes a very long time to expose a 50 foot length of film, taking each frame separately!) you will be able to produce some fine examples of time lapse photography.

# PRACTICAL ELECTRONICS

With effect from the October 1975 issue the price of Practical Electronics will be 35p



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# RIFT ON MARS

The huge Coprates canyon on Mars has puzzled planetary scientists ever since the composite picture of the red planet was made from the many pictures sent back by Mariner 9. The canyon is some 2,700km in length, 150km wide, 6km deep and stretches from the Tharsis plateau toward the east.

Recently two papers have dealt with the subject in a new way. These appeared in "Earth and Planetary

Science Letters".

One of the papers by D. McAdoo and J. Burns, suggests that the cause of the great rent on the surface of the planet is the result of tensions resulting from the outer shell movement. They indicate that in their opinion the stresses were caused when the planet tried to adjust itself to accommodate with the changing position of the poles. The change that is meant here is the wandering position of the poles.

This changing of position would require that the planet's shape also changed as the outer layers sought to adjust the curvature required by the spin on the polar axis. It is this adjustment that could have resulted in the creation of the canyon.

McAdoo and Burns claim that the canyon's proximity to the equator, it's shape, particularly the s-shape, and the east to west alignment all point to "membrane" stresses. This is to some extent supported by other configurations such as the quasicircular formations in the polar regions.

An alternative view is taken by V. E. Courtillet and C. J. Allegre of the Institute de Physique du Globe and M. Mattauer of the Laboratoiire de Géologie Structurale.

They suggest that at the western end of the canyon there may be a junction of three tectonic plates. A "triple junction" as it is known.

One of the plates would be the area of the Tharsis plateau with its numerous volcanoes. Another could be in an area north of the rift and the third could be south between Coprates and the region of Pheonicus Lacus. This is a situation where the transform fault would be similar to the Aden Rift on the Earth.

There will be considerable controversy in this matter in the future for now that two positive suggestions have been made sides will be taken. Certainly there is food for thought in this matter of Mars, for there is a marked difference between its two

hemispheres.

The division in this case, however, is not north and south of the equator but along a line at an angle of 50 degrees to the equator. One hemisphere is heavily cratered, the other is only lightly sprinkled but contains the volcanoes. There is a great contrast between them and this has led to speculation as to the nature of the present state of the evolution of the planet. It is possible, some planetologists say, that Mars could be at a stage where there is one large continent and one ocean basin.

### **NEW COMET**

The period of the first comet discovered by Dr R. West and then by Dr Ikemura of Tokio and Dr L. Kahoutek of Hamburg, is now known to have a period of six years. It is a twelfth magnitude object.

The details of its period has been calculated by Dr R. G. Marsden who says that the comet was at a perihelion on February 26 and was 218 million kilometres from the Earth. It passed within 1.5 million kilometres of Jupiter in March 1972 which probably captured it into near solar orbit at that time. It is now named West-Kohoutek-Ikemura in the catalogues.

### **BIOLOGICAL LINKS WITH SPACE**

From the USSR comes some interesting data about the effects of supernovae on the flora of the Earth. A botanist from Leningrad. Nikolai Lovelius of the Institute of Botany, USSR Academy of Sciences, has demonstrated these effects by using the stump of an 807 years old Juniper tree. The stump came from the slopes of Zerevshan in the Pamir-Altai range of mountains in Turkestan.

This is an area where living organisms are responsive to the slightest

changes in the medium.

Lovelius carefully analysed the volume and distribution of the annular rings and discovered that there were three occasions during the lifetime of the tree when its growth was slowed down. An examination of the cosmic explosions revealed that the time of the tree's slow growth periods coincided with major cosmic events.

The first was at the time of the Tycho Brahe supernova (1572), the second at the time of the Kepler supernova (1604) and that of the third at the time of Cassiopea-A (1700). The growth of the tree was noticeably slower in the first few years after the events. This continued for 15 years before full recovery was reached.

Lovelius does not offer any particular explanation of this new link between outer space and the Earth, but no doubt there are other trees of great age which could be used for study.

### **VENUS PROBE**

The automatic interplanetary space station Venus-9 was launched from the USSR in June. The main aim of this mission is the further exploration of its environment.

The spacecraft will, during its flight along an Earth-Venus trajectory, carry out research into the physical characteristics of the interplantary space including the magnetic fields, solar wind and the ultraviolet radiation.

The Venus-9 is a new type of spacecraft developed for the exploration of the planet. It was injected into an interplanetary orbit from an intermediate Earth orbit. The spacecraft is due to be within the vicinity of the planet in October next.

The flight is being directed from the long range communication centre of the USSR. Many new and novel features are on this vehicle and represent the best of Soviet technology since the launch of the first probe to Venus by the USSR in February 1961.

Venus approaches closer to the Earth than any other planet but this distance constantly varies which complicates launching. It is interesting to make a comparison about the accuracy required, which may be compared to using a rifle to hit a coin at a distance of one kilometre. Although the distance between the Earth and Venus varies between 118 and 85 million kilometres, Venus-9 will have covered around 370 million kilometres by the time of approach.

# **COSMOS SATELLITES**

Two more Earth satellites have been launched by the Soviet Union, Cosmos 741 and Cosmos 742. The orbital elements are:

Cosmos 741
Revolution period: 89 minutes
Apogee: 246km
Perigee: 210km
Orbit inclination: 81.4 degrees

Cosmos 742
Revolution period: 89.8 minutes
Apogee: 375km
Perigee: 189km
Orbit inclination: 62.8 degrees

They were launched on May 29 at

They were launched on May 29 and June 3 respectively.

Cosmos 741 carries a transmitter operating on 19.99MHz in addition to the special equipment for space measurements. A feature on this satellite is a radio system for accurately determining the orbital elements.

# DUGUIAL LC.TESTER

By W.H. DAVIES

T is reasonable to assume that many readers of this magazine possess a transistor tester of one type or another and those who do will readily agree that the expense incurred to effect construction has been repaid ten-fold. Primarily, these testers provided a facility for checking packs of untested and unmarked transistors available from component suppliers at reasonable cost and matching pairs of transistors by comparison of h<sub>fe</sub> measurements.

They also provide a confidence check procedure whereby a transistor can be tested prior to solder-

ing in circuit.

Recent trends in constructional projects indicate the increasing use of logic integrated circuits and a need has arisen for a similar device on which the constructor can test a wide variety of the current

range of i.c.s.

The unit described here does this. It provides a means of testing 14 and 16 pin dual-in-line TTL and DTL types, and light emitting diodes (l.e.d.s) provide a visual indication of the logic states. It is constructed in a box, either metal or plastic, the dimensions of which are determined by the physical size of the switches and associated sockets employed, and whether or not the tester is to be powered by internal batteries or from an external source.

It should be noted, however, that the changeover switches used need not be toggle types as small slider switches are available at low cost and these have an acceptable degree of efficiency. Patch cords are used to obtain the necessary combinations required to permit versatility of operation.

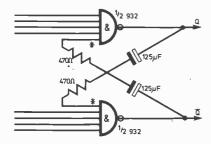
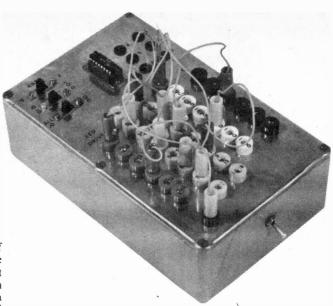


Fig. 1. This basic oscillator circuit can provide two differing frequencies dependent on the values of the two identical capacitors



# COMPONENTS . . .

Resistors

R1 470Ω

 $2 470\Omega$ 

R3 to R8 330 $\Omega$  6 off. All  $\frac{1}{8}$ W 2%

Capacitors

C1, C4 250µF, 10V

C2, C3 125µF 10V

**Integrated Circuits** 

IC1 BP932

IC2 BP936

Diodes

D1 to D6 TIL209 or HP5082

Switches

S1, S2, S3 DPDT slider or toggle, 3 off

SPCO on/off switch

Miscellaneous

Sockets SK1 to SK36, 2mm, 36 off. Plugs to suit. Batteries and holder, 4 off U11 or similar. Case, die-cast or plastic,  $7\frac{1}{2} \times 4\frac{3}{4} \times 2$  in in model.  $3\frac{1}{2} \times 2$  in Veroboard 0-1in matrix, pins, nuts, bolts, spacers, wire, solder, etc.

## THE OSCILLATOR

With the exception of a regulated 5V d.c. power supply source, a pre-requisite for the tester is an oscillator capable of providing the necessary Q and  $\overline{Q}$  square wave logic. This is achieved by employing a standard multivibrator circuit derived from a DTL 930 series BP932 dual 4-input NAND gate, using the expanders with external R and C components as shown at Fig. 1.

The frequency (F) of such a multivibrator is given by the formula:

 $F \simeq \frac{160}{C}$ 

where F is expressed in hertz and C in micro-farads.

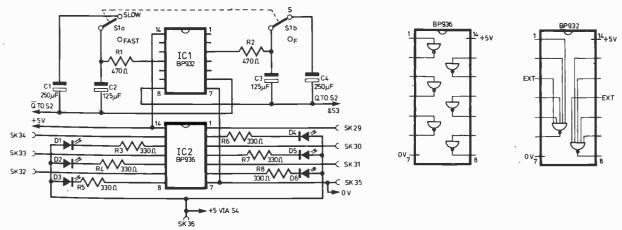


Fig. 2. Circuit diagrams of the oscillator and I.e.d. drive

The oscillator, as constructed, provides a choice of fast (F) or slow (S) modes of operation by switching into circuit one of two values of C,  $125\mu F$  or  $250\mu F$ . The values are chosen so that the two frequencies are 1.28Hz and 0.43Hz respectively. In addition, the Q and  $\overline{Q}$  outputs are switchable so that the available logic may be Q and  $\overline{Q}$  as indicated by the socket labelling on the front panel, reversed Q for  $\overline{Q}$ , or all one (Q or  $\overline{Q}$ ).

The general circuit diagram for the prototype board is shown in Fig. 2. The switch wiring for the  $Q-\overline{Q}$  changeover is shown in Fig. 3.

### L.E.D. DRIVE

A BP936 hex inverter with a 330\Omega load in series with the output of each gate supplies sufficient drive for the l.e.d.s. Gate inputs are via the l.e.d. drive sockets on the front panel. In Fig. 2 the l.e.d.s are shown as on the board with the i.c.s but in fact they, and switch \$1, are mounted on the instrument front panel.

# CONSTRUCTION

The oscillator and l.e.d. drive assembly is mounted on one piece of Veroboard measuring 3½in by 2in, with the copper strips running widthwise as in Fig. 4.

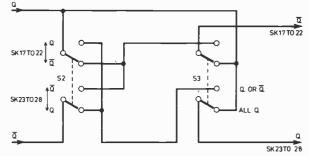


Fig. 3. The Q and  $\overline{\mathbf{Q}}$  output switching for the various combinations of Q and  $\overline{\mathbf{Q}}$  required

Vero pins are used at the various interconnecting points to the front panel and batteries. These provide sufficient anchorage for interconnection wiring and soldering. For the prototype unit the author dismantled an old multisocket connector and used the individual socket elements as female connectors to the Vero pins; this made interconnecting relatively easy.

The switch connections and front panel layout are given at Fig. 5 which shows the wiring diagram for the reverse side of the front panel and the suggested location for the various sockets. The positioning of the oscillator and l.e.d. drive assembly and the battery holder are shown in the accompanying photographs.

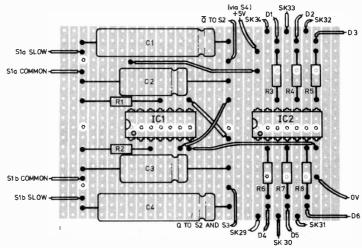


Fig. 4. Component layout and Veroboard cutting details for the oscillator and I.e.d. drive circuitry

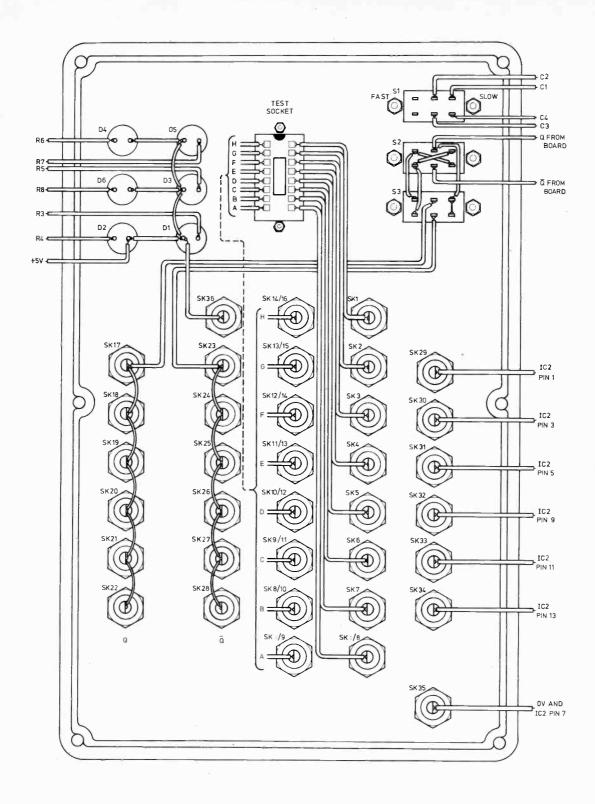
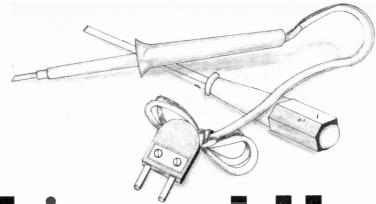


Fig. 5. General layout and wiring of the front panel viewed from the rear showing both switch connections and the wiring of the i.c. test socket patch sockets SK1 to SK14/16



# This could lead to something big.

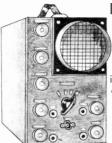
A soldering iron and a screw driver. If you know how to use them, or at least know one end from the other, you know enough to enrol in our unique . home electronics course.

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# Build an oscilloscope.

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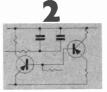
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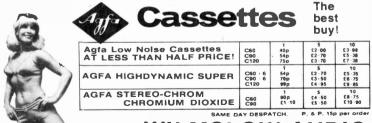
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#### PATCH CORDS

A minimum of 18 patch cords is desirable but not essential. However, at least two of these should be of the triple cord variety, i.e. three cords each having an individual banana plug at one end, but terminated into a common banana plug at the other.

#### **OPERATION**

Numbering the l.e.d. inputs 1 to 6 on the front panel from top to bottom, D4 becomes 1, D5 is 2, D6 is 3 and so on.

On completion of the unit it is suggested that the Q and  $\overline{Q}$  sockets be tested for correct logic by patching these sockets direct to the l.e.d. drive socket. If the oscillator is functioning correctly the l.e.d.s will follow the logic and flash on and off.

With the patch cords still in this position proceed to test the FAST and SLOW modes of operation by operating the RATE switch. Then check the logic reversal facility by operating the  $Q-\overline{Q}$  changeover switch.

Finally check the ALL Q facility by operating  $Q/\overline{Q}$ —ALL Q switch. If the unit is functioning correctly we may now proceed to the next stage, that of checking known i.c.s and some examples are given.

#### 7404 HEX INVERTER

To test a 7404 TTL hex inverter, patch socket SK14/16 to 5V, SK36, SK7 to 0V SK35, socket SK1 to a Q socket, socket SK2 to l.e.d. drive SK29, and l.e.d. drive SK30 to a  $\overline{Q}$  socket. Now insert the 7404 into the holder ensuring that it is correctly positioned for 14 pin d.i.l. operation and switch on power supply.

Light emitting diodes D4 and D5 should now flash on and off simultaneously, indicating that the inverter gate is functioning correctly. Repeat the procedure for the remaining five gates, bearing in mind that gate inputs are via sockets SK3, 5, 9, 11 and 13 and gate outputs via sockets SK4, 6, 8, 10 and 12 respectively.

#### 7490 DECADE DIVIDER

Patch socket SK5 to the +5V socket SK36, and socket SK10 to the 0V socket SK35 using a triple lead patch cord. Connect the remaining two leads of this patch cord—one to socket SK2 and one to socket SK6.

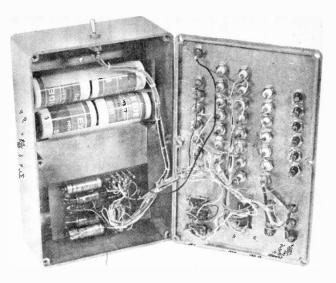
Patch socket SK14 to a Q socket and socket SK12 to socket SK1. Patch socket SK11 to l.e.d. drive SK29, and l.e.d. drive SK30 to a Q socket.

Insert the 7490 ensuring that it is correctly positioned. Switch on power supply.

Diode D4 should light once for every ten pulses of diode D5 indicating that the correct frequency division has taken place.

## 7413 DUAL 4-INPUT NAND (SCHMITT TRIGGER)

Patch socket SK14/16 to the 5V socket SK36. Patch socket SK7 to the 0V socket SK35. Patch SK1, 2, 4 and 5 to a Q socket and socket SK6 to l.e.d. drive SK29.



Patch l.e.d. drive SK30 to a  $\overline{Q}$  socket, and insert the 7413 ensuring that it is correctly positioned. Switch on power supply.

Diodes D4 and D5 should now flash on and off simultaneously indicating that the first 4-input NAND gate is functioning correctly. Repeat the procedure for the second gate substituting socket numbers SK9, 10, 12 and 13 for sockets SK1, 2, 4 and 5, and SK8 for SK6.

#### 7492 DIVIDE-BY-TWELVE

Patch SK5 to the 5V socket SK36. Patch socket SK10 to the 0V socket SK35 using a triple lead patch cord. Connect one of the remaining two leads to socket SK6.

Patch socket SK1 to socket SK12 and l.e.d. drive SK29 to a Q socket. Patch socket SK8 to l.e.d. drive SK30, connect socket SK14 to a Q socket and insert the 7492.

Switch on the power supply and diode D5 will illuminate after six pulses of D4 and remain lit for the next six pulses, indicating that the correct frequency division has taken place.

Sufficient examples have now been given for the constructor to familiarise himself with the capability and operation of the tester and no difficulty should be experienced when testing other devices provided pin connection data and truth tables are available. Some devices such as retriggerable monostables may, in some circumstances, need two external components in the form of an R and C, but these may easily be connected into circuit for test purposes by wedging the ends of the components into the appropriate sockets with the aid of the banana plugs.

#### MATERIALS

The two DTL i.c.s BP932 and BP936 used in the oscillator and l.e.d. drive assembly are obtainable from Messrs BI-PAK, or A. W. Marshall and Sons. The case is a normal die-cast box housing all components and the i.c. socket is a normal 16-pin holder with a bezel made up to suit. Of course it may be adhered to the case with Araldite or the like.

# SEMICONDUCTOR UPDATE...

By R.W. COLES

#### LONG TIME, SMALL C

Last month I reported on the ZN1040E count and display circuit. an exciting example of the Ferranti C.D.I. process which is now being used in a number of interesting standard circuits, complementing the custom designs which the process has been associated with in the past. Another example of the power of standard C.D.I. circuits is provided by the new ZN1034E precision timer, which fills a big gap in the range of timing circuitry available to the circuit designer.

Until this new i.c. came along, the standard way of producing a time delay was to use a monostable circuit of some kind, a prime example being the popular and useful 555 chip which has appeared in many

amateur designs.

Timers such as the 555 are limited in the length of time delay attainable by the practical limitations of the large C and R components required for long time periods. In the case of the 555 for example, the maximum delay that can be realistically achieved is about ten minutes.

To produce longer delays, some form of counter circuit has to be added which will require resetting circuitry, and can consume considerable power, two reasons why this is a messy solution not suited to simple constructions which need

a long delay.

The ZN1034E is the simple answer for producing long delays because it contains in one 14-pin d.i.l. package, a very stable oscillator and a 12-stage binary counter which gives a division of 4095. The time period required is set with the usual C and R components external to the chip, but now the calculated time constant (C times R) can be multiplied by 4095 to derive the overall time delay produced. As a striking example, a resistor of 1 megohm and a capacitor of 1µF will give a delay well in excess of one hour, the exact value being trimmable by means of a calibration resistance which could be a pot.

The chip can be operated from a TTL 5V supply, or, by standard using the internal regulator circuit, a supply of from 6 to 450V can be employed. Typical consumption is a miserly 5mA which makes battery operation quite feasible, and I think this is a device that we will be seeing more of in these pages.

#### A TASTE OF TOMORROW

An indication of just how much sophisticated electronics will soon

be affecting our day-to-day lives is provided by a data sheet on two new devices from Texas Instruments.

Even though we are unlikely to see these components offered for sale to amateurs in the forseeable future, the mind-boggling implications of their eventual widespread use by industry is sufficient justification for their introduction on this page. I am referring to the TMS1000NC and the TMS1200NC one chip microprocessor, devices which contain a complete digital computer including program store, data store, and a versatile arithmetic unit.

The unique feature of these particular devices is that they can be programmed to suit the user's requirements at the manufacturing stage, so that unlike, for example, the dedicated calculator chips we hear so much about these days, the TMS1000 or 1200 will soon be found in applications as diverse as cash registers, motor vehicles, vending machines, and electric cookers, in addition to the more traditional data-processing areas such as computer terminals and printers.

There are, of course, other chips around which can be used in these types of application, but in general they require a lot of back-up circuitry which means that they are expensive to use and unsuitable for "consumer" goods.

The new Texas devices can stand entirely alone in many circuits, and can be programmed to accept and display data in a wide variety of forms. The chameleon like qualities of these components are provided by the ability to build into the chip 1024 different programme steps which will decide whether it becomes a car nerve centre or a cash register brain, a potentially very cheap way of equipment design.-Roll on 19841 1 1

#### BRIGHTER THAN BRIGHT

Nobody can deny that l.e.d.s are amonast the most useful devices to appear in recent years, but despite all their advantages, a few problems have remained.

One of the first serious criticisms of these devices was directed against the fact that you could have any colour, so long as it was red; a problem that semiconductor manufacturers have been acutely conscious of, and one which they have overcome by developing exotic new compounds which have added green and yellow light to their repertoire.

Another criticism sometimes levelled at l.e.d.s is that their luminous efficiency is quite low, which means that you don't get much light power for the electric power you drive them with. Of course, this means little when plenty of drive power is available, but in today's energy conscious world, the era of microwatt CMOS logic circuits, the power wasted in displays can be a thorn in the side of the designers of portable equipment.

In the forefront of technology as Hewlett Packard introduced a new range 'Brighter-Than-Bright' Le.d.swhich are many times brighter than their predecessors while consuming the same power. The new range is available in the three colours (with no premium to be paid for green or yellow), and two lens styles giving a wide beam (diffused), or a narrow beam (point source).

Hewlett Packard codes for the new discrete l.e.d.s are as follows: 5082-4658 (red), 5082-4558 (yellow), 5082-4958 (green) for the point and 5082-4655 (red), 5082-4555 (yellow), 5082-4955 (green) for the diffused type.

The secret of the new devices is a new construction technique which uses a transparent gallium-phosphide I.e.d. chip backed up by an integral reflector, resulting in all the available light being directed exactly where it is required.

Those who are fed up with peering their 'Dimmer than Dim' calculator displays can take heart from this new technology which promises the chance of a brighter display while retaining long battery

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# WALC. Ainslie By A.C. Ainslie Control By

THIS second part deals with construction and testing of this versatile oscilloscope adjunct.

#### CONSTRUCTION

The circuit is built on three p.c.b.s—one for the p.s.u. and one for the D/A converter and the third main board for all the rest.

The print drawings and component layouts are given clearly in the following figures.

The p.c.b.s can all be made using the etchant resist pens now freely available. The larger one is a rather tedious job but can be completed in a few hours if the layout is drawn in carefully and the position of the various chips is marked.

Veroboard or Matrixboard could doubtless be used for this project but would entail some complex cross linking.

The main board has wire links for almost all of the power supply wiring and the parallel inputs to IC1 and IC4 are also formed using cross links.

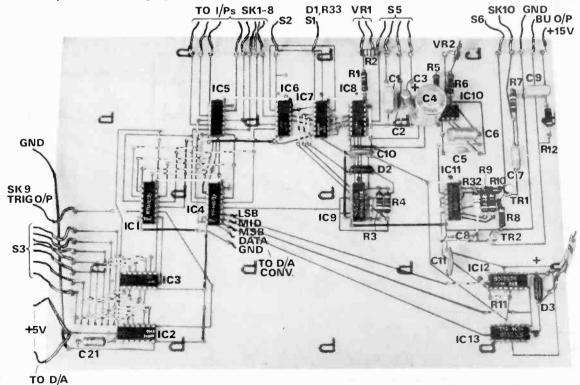


Fig. 6. The main board showing component layout and interconnections to other components and boards

BOARD

Similarly the wiring for the decimal/binary converter which feeds IC1 is also cross wired. The two inputs to the 74121 involve a crossover but this is made clear on the layout. The only other cross wiring is where necessity demands. Obviously if double sided board could be used this would be a considerable advantage.

Potentiometer VR2 was originally intended to be a back panel preset but a skeleton soldered in position as in Fig. 6 is just as convenient. S1 is ganged with VR1 for convenience of operation.

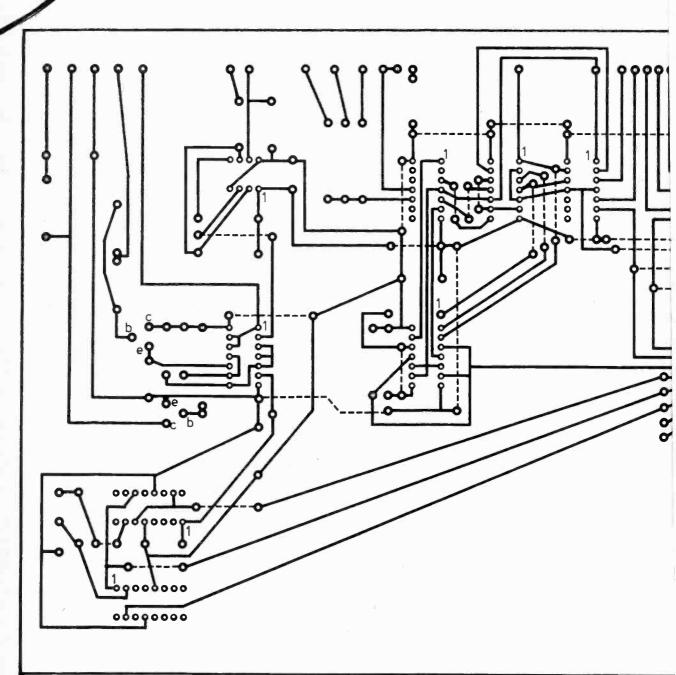
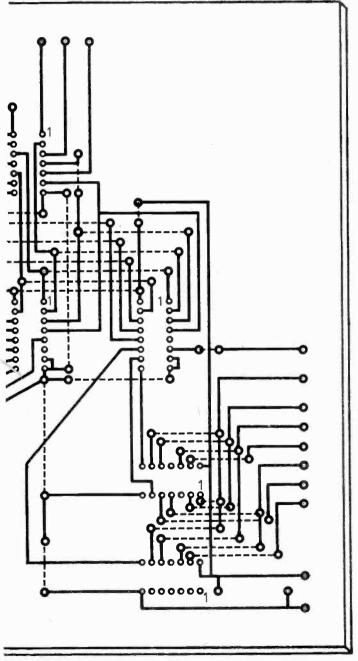


Fig. 7. The printed circuit master for the main board of Fig. 6

Small connecting pins are used for all external connections, including those to the D/A board.

The D/A board is quite simple, there being only one wire link shown in Fig. 8. The earth tracks should be kept as wide as possible. The Gnd. point next to the output is the only connection to the output lead braiding. The D/A board is intended to fit on top of the main p.c.b. in the vacant area. A number of short lengths of single flex twisted together constitute the input and ground connections.



#### **POWER SUPPLY**

The power supply board (Fig. 9) is also quite easy to assemble. FS2 is mounted directly onto the board with 1½in p.c.b. fuse holder. Alternatively, it could be made externally accessible by mounting a fuse holder on the rear panel. However, as the fuse should only blow under fault conditions, this is not really necessary.

The original board was designed so that a separate 24V-0-24V winding could supply a higher voltage to TR2. However, the smoothed 15V line proved adequate and so the points A and B are linked across. Otherwise, should a more vigorous bright-up be found necessary, the link can be removed, a 100 µF coupled between A and B and two diodes connected at D and E—driven from a centre tapped winding of up to 30V for a 2N706 bright-up transistor. Higher voltages can be used with other transistors but a Z axis amplifier in the scope would be a better alternative.

TR3 is mounted on a heat sink of about 3 to 4in square. The power dissipated is about 3W and so the device runs well within its rating. IC15 has a relatively easy task and just gets warm. D6 does nothing until the supply tries to exceed 5·1V when it starts conducting, blowing FS2. So under normal conditions D6 should be cold. A warm device would indicate that the 5V line was adjusted to be on the high side.

#### CABINET

The prototype was built in a commercial cabinet of dimensions about  $8 \times 2\frac{1}{2} \times 13$ in. The layout can be clearly seen from the photographs.

In order to ease selection of the trigger channel in the prototype, eight small filament lamps were wired from +5V to S3. Thus rotating S3 caused the light appropriate to the trigger channel selected to light. The lights were placed above their respective input sockets. L.e.d.s could also be used with a 470\$\Omega\$ in series but they are expensive.

The prototype front panel was lettered with "Letraset" and finished with polyurethane spray. Chassis mounting BNC sockets were used for all the inputs and the "B.U." and "Trig" outputs. These sockets are grounded by the metal work which is grounded to the electronics by a short lead from the power supply ground to a solder tag bolted to the chassis. The ground connection to S3 is made to the same point as are all the other ground connections.

#### WIRING

There is quite a lot of front panel wiring to complete and this is best done in the form of a harness. The lead to the B.U. socket radiates spikes as does the lead from the scope timebase gate input which is mounted on the back panel. These two leads are therefore run around the perimeter of the instrument, out of harm's way.

The output lead from the D/A is a 36in length of 500 cable terminated in a BNC plug to plug directly into the scope Y channel. This lead is grounded only at the point adjacent to the D/A output. Originally a BNC socket was chassis mounted on the front but was susceptible to earth return spikes which blur the display.

# D/A BOARD DETAILS

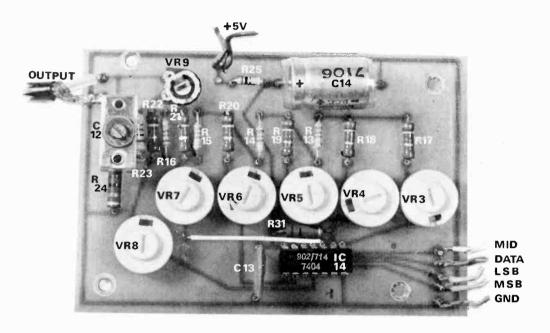
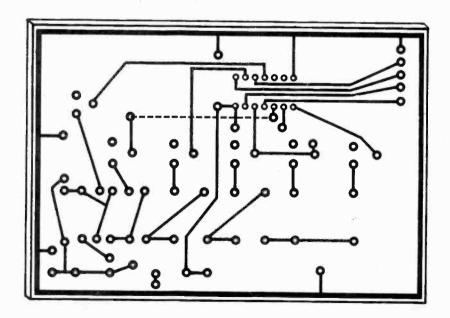


Fig. 8. The D/A board component layout (above) and p.c.b. master (below)



# POWER SUPPLY UNIT

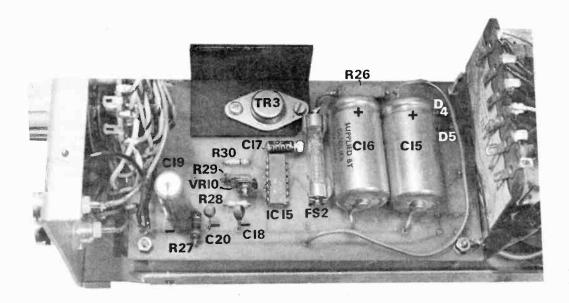
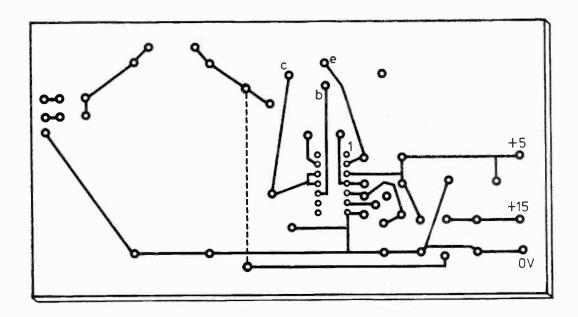


Fig. 9. Power supply layout (above) and associated p.c.b. master below



The p.c.b.s are mounted on 4in stand-offs so that the metal work forms a reasonable ground plane for the electronics.

#### TESTING AND SETTING UP

Initially VR10 is set to maximum resistance. The mains is applied and the 5V rail is measured (it is pertinent to replace the logic load by a resistor of 15\Omega and 10W rating). Decreasing VR10 should increase the supply towards 5V. With the supply set to precisely 5V, D6 should not get even slightly warm. Setting the supply higher will cause D6 to overheat and probably blow the fuse.

With the logic connected, pin 3 of IC10 should have a roughly square wave of a few tens of kHz. VR2 is decreased until the output wave train starts to jitter. At this point the steps of the staircase will

be non uniform.

The correct setting of VR2 is just prior to the onset of jitter.

With S6 set to "CHOP" and S1 open, a staircase should be present at the D/A output if S4 is set to the 8-trace position.

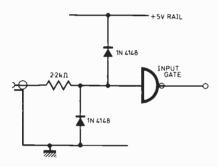


Fig. 10. Input circuit designed to reduce the risk of input overload

#### LINEARITY

The ladder is now set up to give a linear output. With VR3 and VR6 at mid position, VR7 is adjusted to make the space between traces 2 and 3 the same as between 1 and 2. Similarly VR8 adjusts the spacing of lines 4 and 5. VR9 is now set to give one step per unit with the Y input to the scope set to a convenient step.

Grounding all inputs should cause all the traces to drop by 0.75 units. VR3, VR4 and VR5 are now adjusted to ensure that the ladder is again linear.

With the inputs floating again the other presets can be used to trim up the linearity. The settings are somewhat interdependent but very good linearity can be achieved without much difficulty. C12 adjusts overall sharpness of the step transitions.

Connecting the B.U. lead to the Z axis input should cause the actual transitions on the display to disappear. The properly triggered display should now be a series of short clean horizontal lines.

Applying a logic signal to channel 8 and switching S3 to select CH8 as trigger source should cause the channel 8 signal to appear at "Trig Output", SK9.

Connecting this lead to the scope external trigger select should then enable the lower trace to display the CH8 signal, all other seven traces being "high". Reducing the scope sweep speed and switching to "ALT" with the scope sweep gate connected to SK10 should cause the traces to be displayed sequentially, from 1 to 8. Switching S4 should allow 8, 4 or 2 traces to be displayed. Even in the 4 or 2 trace mode, channels 1, 2, 3, etc. can be used as trigger sources.

#### **PULSE STRETCH**

The next step is to check the pulse stretch operation. If a 50ns wide pulse is applied to CH8 input with a 1kHz p.r.f. then the pulse will not be visible at 100µs/cm sweep speed. If the trace is low then the applied pulse must be positive going. S2 is therefore left open. S1 is closed and D1 should light. Rotating VR1 in conjunction with S5 should show a positive pulse, of length dependent on the setting of these controls.

A negative going pulse (the trace appears to be high) is stretched with S2 closed and will then appear as a negative going pulse of width determined by VR1/S5.

In operation the unit is very straightforward to use and has proved to be of immense value in fault

finding and debugging of logic systems.

The Logic Trace Multiplier is intended for use with TTL systems and must not be used on any other system without providing an interface. To reduce the possibility of damaging the input i.c.s due to input overload the circuit in Fig. 10 may be used, with a slight loss of speed. The components can be conveniently mounted on the BNC sockets.

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#### SECURITY ALARM

A new modular burglar alarm system suitable for home security protection, which, it is claimed, can be installed anywhere without the need for professional skill is now being marketed by Euroswitch Ltd. The system, called the Euralarm I, comprises a control box, external bell box, a 6in diameter bell, six magnetically operated reed switches (door contacts), two pressure pads and wire, insulated staples and connector blocks.

The control box, protected by an anti-tamper microswitch, is normally mains operated, but a battery can be placed in the box to give a standby supply if mains failure occurs. The control circuit provides for single and double pole systems, has a facility for testing the alarm circuit and has an open circuit facility which allows the pressure pads to be incorporated. Externally, the control box has a key switch, mains on indicator light, a mains fuse access cap, a push to test switch and a green systems operating light.

The alarm is easily installed by simply placing the units in the required positions and running the wire between contacts, control box and bell. The bell box also has an anti-tamper microswitch fitted.

The Euralarm 1 costs £45 and can be obtained direct from Euroswitch Ltd., 121 Mildmay Road, London, N.1.

#### INSTRUMENT CASES

With economy in mind, West Hyde Developments have just introduced a range of equipment cabinets.

Known as the Contil Swift range these cases feature the use of the strengthening extrusions as semi-conductor heat sinks and mounting supports for such components as transformers. By mounting these components directly to the case extrusions, this eliminates the use of mounting brackets and the need for separate heat sinks.

Full particulars of the complete range of Contil Swift cases can be obtained from West Hyde Developments Ltd., Ryefield Crescent, Northwood Hills, Middlesex, HA6 1NN.

#### BEGINNERS CONSTRUCTION KIT

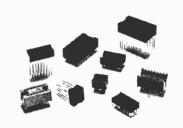
One of the latest products from Heath (Gloucester) Ltd. is an experimental, educational construction kit, type IK-18A

type JK-18A.
This kit er

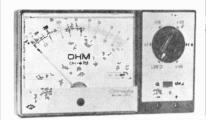
This kit enables the beginner to construct up to 35 different circuits and, at the same time, learn how each component functions. Spring terminal connections eliminate any soldering and all circuits are battery powered.

# MARKET

Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned. All quoted prices are those at the time of going to press.



Dual in-line sockets from Dieter Assmann



OH470 ohmmeter from Chinaglia



#### Euroswitch alarm kit

Ranging from a 4 transistor radio to a rain alarm, all projects begin with an explanation of what the selected circuit does and how it is used.

A large schematic diagram is furnished for each experiment and includes a simple but complete explanation of what each part does. Each component is identified by a drawing and its schematic equivalent as used in circuit diagrams so that the constructor learns to read schematics while experimenting.

schematics while experimenting.
Full details of projects and price list for the Heath JK-18A kit is available from Heath (Gloucester)
Ltd, Bristol Road, Gloucester, GL2
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#### RESISTANCE METER

Designed for resistance measurements over a wide range with an accuracy of 2.5 per cent, Chinaglia have developed the OH470 multirange ohmmeter.

Using a  $40\mu A$  moving coil movement the meter has six switched ranges covering 0·1 ohm to 50 Megohms. The meter is powered by two 1·5V internal batteries, includes test probes and a protective plastics case.

Further information on the complete range of Chinaglia meters is available from Chinaglia, 19 Mulberry Walk, London, SW3 6DZ.

#### **DUAL IN-LINE SOCKETS**

A comprehensive range of sockets for dual in-line packages are now being marketed by Dieter Assmann Electronics Ltd.

Socket for both in-line and staggered configurations of up to 40-pins can be supplied from stock for soldering directly into printed circuit boards, for use with free wiring or for wrapped wire connection.

For applications in which low profile is important a range of 8-pin, 14-pin and 16-pin sockets, type ZK and ZKF, having heights of 5mm and 3-9mm above the printed circuit board, are available.

Information on the complete series of sockets available can be obtained from Dieter Assmann Electronics Ltd., Victoria Works, Water Lane, Watford, Herts. WD1 2NW.

#### SPECIAL OFFER

In an effort to fulfil the increasing demands of the electronics enthusiast for competitively-priced, quality components, A. Marshall (London) Ltd., established since 1953, have been opening retail shops in Glasgow, Paris and Bristol carrying their new extended range of components. The 1975 Marshall's catalogue now carries a range of over 3,000 types of semi-conductors as well as an extended range of passive components and accessories.

To mark the opening of their new components shop in Bristol, towards the end of August, A. Marshall & Son Ltd. are making a special offer to all P.E. readers this month.

Readers of P.E. can purchase first class branded CMOS and TTL devices at special discount prices. This offer is for one month only.

For full details of devices available and prices, readers should turn to the A. Marshall advertisement in this

#### **APOLOGY**

Due to production difficulties and recent VAT changes Henry's Radio would like to apologise to readers of P.E. for the delay in dispatching copies of their 1975 Components Catalogue.

Copies are now being dispatched and readers who have ordered copies should now be receiving them.

The basic philosophy behind the cabinet design is to allow maximum wiring to be carried out from the rear, simplifying the interconnections between the keyswitch and the p.c.b.s. This is a considerable improvement over the prototype arrangement (see page 389 of the May issue) in which wiring was carried out from the top, thus making it impossible to remove a single board with ease, and reduces the wiring complexity.

#### KEYBOARD POSITION

The keyboard spans a gap between the front baseboard and the centre base spar, allowing immediate access to the keyswitch, and a further 4in gap exists between the spar and the rear panel of the cabinet, revealing the terminal pin edge of all p.c.b.s (Fig. 5.1).

The two base members in combination with the rear panel provide the main rigidity in the cabinet, further strengthened by the p.c.b. supports and keyboard when fitted and fix to the two shaped end panels. The p.c.b. supports have been designed to hold the p.c.b.s. provide fixing points for the front panel, and support to the top panel. It should be noticed that the p.c.b. slots are spaced differently on each support. The bottom retainer strips prevent the p.c.b.s falling to the bottom of the cabinet. A bottom cover panel made of pegboard covers the wiring but allows ventilation of the cabinet.

Magnetic catches have been used on the prototype to retain the top panel.

#### ASSEMBLY OF THE CABINET

The front baseboard, centre base spar, rear and end panels should first be assembled using joints to match the skill of the constructor. The p.c.b. supports should then be fixed to the rear panel and centre base spar, and the p.c.b. retainers glued in position.

At this point the keyboard should be fixed into the cabinet inserting from the front. Large washers are used to trap the keyboard channel edging to both base bars ensuring that the keyboard is central by using the end cheeks as guides. When the keyboard is fixed, the keyboard trims can be fitted, followed by the end cheeks which are screwed down to the keyboard channel after drilling suitable holes in the channel. The top panel support, front panel support and front panel can now be fitted, followed by the rear p.s.u. support. If the unit has been constructed accurately the top and bottom cover panels should easily fit into place, and the magnetic catches can then be fixed.

#### KEYSWITCH ASSEMBLY

The keyswitch assembly consists of a printed circuit board to which gold plated springy wires are soldered and seven plastic blocks are fixed to support two gold plated rods (Fig. 5.2).

Before assembling the kit of parts together, the p.c.b. should be placed on the underside of the keyboard, and used as a template for drilling the six 3 in holes required to take the self-tapping screws



By A.J. BOOTHMAN B.Sc.



which fix the switch assembly to the frame. The front edge of the switch assembly should be placed approximately  $\frac{1}{4}$ in from the keyboard actuators, and should be centralised over its length with respect to the actuators. When holes have been drilled, and the p.c.b. removed from the frame, the keyswitch is assembled as follows:

- 1. Slide the seven rod supports onto one rod.
- 2. Insert the second rod through the remaining rod holes, and space the supports roughly opposite the holes which have been countersunk from the non-copper side of the p.c.b.
- 3. Fix the rod supports to the copper side of the p.c.b, using the countersunk screws, nuts and locking washers. The track is formed to indicate the correct positions.
- Cut the gold plated wire into 61 lengths of l<sup>3</sup>/<sub>4</sub>in, and tin approximately <sup>1</sup>/<sub>4</sub>in of one end of each wire.

- 5. Pass each wire between the two rods and solder the tinned end to a main pad. About ½in should be left to hang over the front edge of the p.c.b. The wire should be pressed flat to the p.c.b. close to the main pad using a matchstick, and the soldering iron applied to the wire and land for as short a time as possible.
- The assembly can now be fixed to the underside of the keyboard using the self-tapping screws with two washers under each screw head.
- 7. The next operation is to adjust the bend of each wire to ensure that it seats satisfactorily against each rod in turn, whilst gently resting on the actuator in the non-depressed state. An attempt should be made to achieve a gentle curve in the wire commencing on the land side of the rod in order to prevent too much localised stress where the wire presses against the top rod.

The touch resistors are soldered to the pads provided, one end to the common touch rail, and the other to the land which also carries the gold-plated springy wire. Top and bottom octaves should be  $1.8 \,\mathrm{k}\Omega$ , and the middle three octaves  $1.5 \,\mathrm{k}\Omega$ . The resistor leads should be bent back under the resistor and cut short before soldering to the board, which then allows the component to be fixed within the small space available.

#### INTERWIRING PROCEDURE

The p.c.b.s should be inserted into the slots with components facing forward and the top panel fitted.

#### nd out

Felt Strip

Magnetic Catches 2 off

#### ——— CUTTING LIST

CABINET	
Front Base Board	36≩in × 4in × ≨in
Centre Base Spar	36≩in × 2┧in × ┧in
Rear Panel	38in × 4⅓in × ∦in
End Panels (2 off)	14in × 4ీin × ẫin
Front Keyboard Trim	36≩in × 1┧in × გin
Front (fascia) Panel	36¾in × 1⅓in × ⅓in
Top Panel	36≩in × 7in × ⅓in
Bottom Cover Panel	
(Pegboard)	36¾in × 10¼in × åin
P.c.b. Supports—shaped (4 off)	6≩in × 3↓in × ≩in
P.c.b. Bottom Retainers (4 off)	$4in \times \frac{7}{8}in \times \frac{1}{8}in$
Top Panel Support	
(p.s.u. end)	4≩in × 1in × ↓in
Front Panel Support	
(p.s.u. end)	2ain × ½in × ½in
End Cheeks (2 off)	$7in \times 1\frac{1}{2}in \times \frac{7}{8}in$
Rear p.s.u. Support	11in × ½in × ½in

LEGS Top Spar Bottom Spar Legs	(2 off) (2 off) (4 off)	12½in × 1½in × ¾in 14in × 1¾in × 1in Chrome plated tubing 24in × ¾in dia.
Bottom Tube		Chrome plated tubing 26in × ¾in dia

36∄in × ⅓in × ⅓in

PEDALS	
Rear Block	$6\frac{1}{2}$ in $\times$ 2in $\times$ 1in
Foot Pedals (2 off)	5i̇̃n × 1in × ∦in
Base	7±in × 7in × ±in
Springs and hinges	_

STOOLEnd Panels (2 off) $19in \times 16in \times \frac{2}{3}in$ Top/Bottom Panels (2 off) $20in \times 14in \times \frac{2}{3}in$ Side Panels (2 off) $20in \times 6\frac{1}{2}in \times \frac{1}{2}in$ 

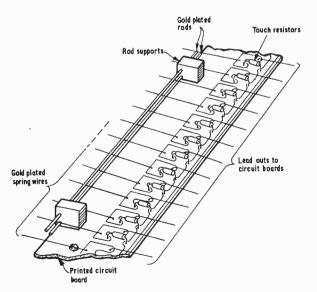


Fig. 5.2. Keyswitch assembly is screwed to the underside of the keyboard so that the actuators press against the spring wires

The cabinet should be turned upside down to reveal the back edge of the p.c.b.s (which have been effectively inserted upside down) and the keyswitch previously fitted as above. Wiring should be carried out as follows:

- 1. A group of three wires (suggest red, orange, yellow) should be soldered to the  $C_1$ ,  $C_1\#$ , and  $D_1$  pads on the keyswitch and threaded through a hole drilled through in the centre base spar. The hole is opposite the correct pins on the three Envelope Boards, and should be connected in the same order.
- A group of five wires (red, orange, yellow, green, blue) should be soldered to the E<sup>h</sup><sub>1</sub>, E<sub>1</sub>, F<sub>2</sub>, E<sub>1</sub> #, G<sub>1</sub> pads on the keyswitch and threaded through another drilled hole.
   This hole is opposite the pins in five Envelope Boards, and the wires should be soldered in the same order.
- The operation should be repeated for A<sub>1</sub><sup>h</sup>, A<sub>1</sub>, B<sup>h</sup><sub>1</sub>, B<sub>1</sub> (red, orange, yellow, green), passing through the A<sup>h</sup><sub>1</sub>-B<sub>1</sub> hole in the centre base spar.

- A further group of three wires (red, orange, yellow) should be soldered to the C<sub>2</sub>, C<sub>2</sub> #, and D<sub>2</sub> pads, threaded through the hole marked C<sub>2</sub>-D<sub>2</sub> and soldered to the relevant boards.
- 5. If this operation is repeated for 60 keyswitch pads (excluding C<sub>a</sub>), each Envelope Board uses a single colour for five connections and is easily seen to be correct. The C<sub>a</sub> connection is made to the Tone Generator Board, which has been inserted upside down.

### TONE GENERATOR BOARD TUNING LOCATION

Since the Tone Generator Board has been inverted, the tuning potentiometer is not accessible. Dependent on the style of potentiometer used it may be possible to reposition it in the back of the p.c.b. in such a position that it is easily available from the top. Care should be taken to avoid shorting the copper track when an adjustment of juning is being carried out.

#### INTERWIRING BETWEEN BOARDS

The next operation is to link the Envelope Boards for the outputs,  $\pm 5V$ , zero volt, and sustain rails. This is best achieved using bare single strand wire pre-cut to the required length. The G connections associated with each input (L, M and H) on the Voice Board (Fig. 4.8) are redundant and should be ignored. Since the pins are double-ended, one half protruding from each side of the board, links can be soldered between the pins which can be easily undone should it be necessary to remove a single board at a later date. When all the boards are interlinked the interwiring of the three rear boards should be carried out as shown, leaving flying leads for connection to the  $\pm 5V$  and zero volt connections on the Power Supply Unit.

At this point connections should be made to the two keyswitch rod ends. Black nearest the keyboard and red furthest away from the keyboard, and a third colour of wire connected to the common touch resistor track on the keyswitch. The three wires should be threaded through the relevant hole in the centre base spar, and left long enough for later connection to the power supply.

Still referring to the wiring diagram, connections should be made between the C Envelope Board and the Voice/Preamp Board—i.e. L. M and H outputs, and the common sustain connection between Voice/Preamp and the F Envelope Board.

Zero volt, +5V, and high output connections should then be made to the Tone Generator Board from the  $\Lambda^b$  Board. The semitone outputs from the Tone Generator Board are wired to  $\Lambda^b$ ,  $\Lambda$ ,  $B^b$ , and B envelope signal input pins, whilst the remaining eight leads are passed through the centre base spar and distributed to the relevant Envelope Boards via the holes in the spar marked  $E^b$ -G and C-D signal input. Colour coding of these wires is desirable, and this can be arranged in a group of three (C-D) and a group of the five ( $E^b$ -G).

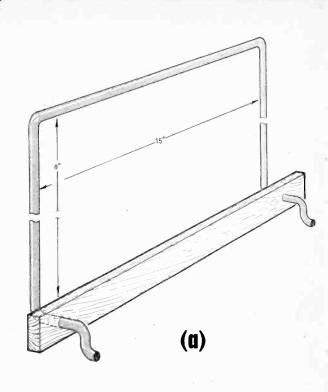
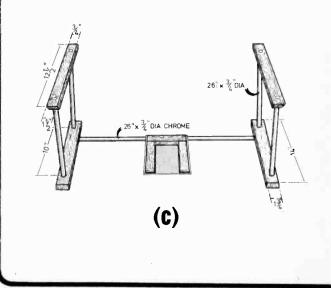
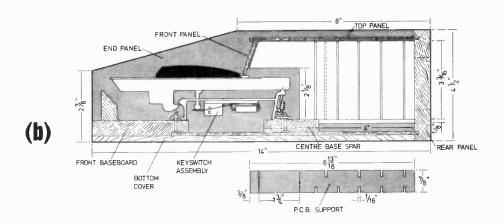
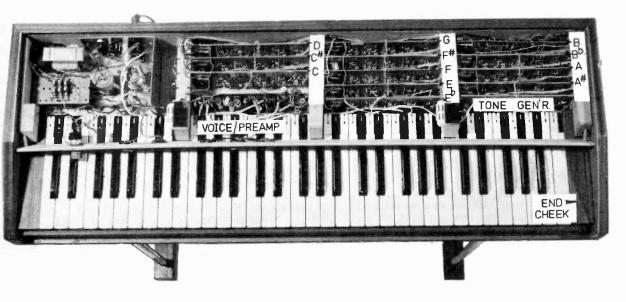
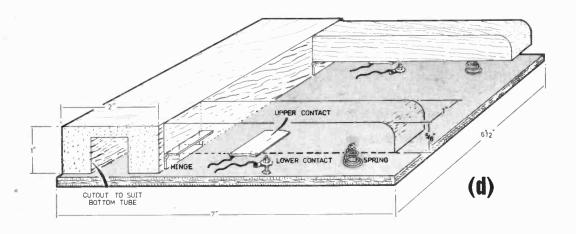


Fig. 5.1 (a). Music stand; (b) cross section of "Joanna" cabinet assembly; (c) leg assembly; (d) soft and sustain pedal details









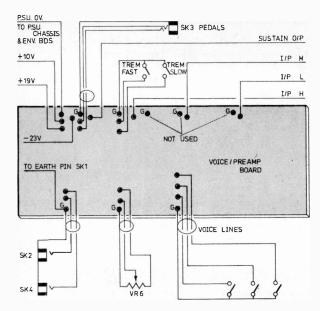


Fig. 5.3. Connection to Voice/Preamp Board including ground connections to minimise hum

#### WIRING FROM THE TOP

Before commencing top wiring the level control potentiometer, mains neon, and five control switches should be fixed in their respective positions on the front panel. Wiring can then be carried out in the following order:

- Long leads (colour coded) should be soldered to the +19V (+V), and -23V (-V) pins on the Tone Generator Board, passed between the p.c.b. supports and keyswitch ready to connect to the Power Supply.
- The four supply leads should be connected to the Voice/Preamp Board with similar coding for +19V and -23V as above.
- 3. All connections can now be made to the Power Supply Unit. The +10V supply (V/2) is obtained from the junction of C3 and IC1 in Fig. 2.1.
- 4. Coded interconnections should be made between the Voice/Preamp Board and SK2 and 4 and SK3 on the p.s.u. chassis, preferably using 2-core coaxial lead in both cases.
- 5. Coded interconnections should be made to the tremolo switches and voice switches, using coaxial lead for the latter link.
- A coaxial lead should be used to link the Voice/Preamp Board and the level potentiometer.
- Mains switch and neon connections should be made to the Power Supply Unit.

The various links between the Voice/Preamp Board and switches and sockets are illustrated in greater detail in Fig. 5.3, along with the ground interconnections to avoid any hum due to earth loops.

#### LEG ASSEMBLY

Details of the legs are shown in Fig. 5.1 and are constructed from four preferably hardwood blocks shaped and dimensioned as indicated, interconnected with chromium plated ‡in diameter tubing. The legs are screwed to the Cabinet, into both front and centre base spars.

#### **PEDALS**

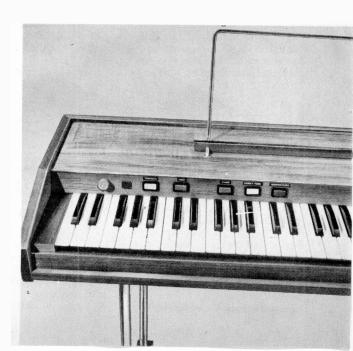
Pedal details are also given in Fig. 5.1. Again hardwood is used throughout.

This simple assembly, whilst fairly fragile, can stand up to sensible use, and is designed to give a neat appearance rather than for maximum strength. The rear block should be channelled or drilled to accept the bottom tube as a retainer. The simple switches are made by gluing phosphor bronze strip under the pedals to contact with the ends of screws recessed into the base. Compression springs should be held in position by indentations in the base and pedals.

#### **POWER AMPLIFIER**

The electronic piano is a device which is generally improved by high peak power capability, one commercial manufacturer incorporated a 150W peak power rated system into his design. Completely satisfactory results have however been obtained from the system quoted above, and it is possible that a rating between 15 and 25W (sine wave) could be acceptable.

One amplifier parameter only seems to have greater than normal importance in this application and that is intermodulation distortion. This is the effect where the mixing of two notes, say at a third interval, produces a strong difference frequency signal which is detectable as a non musically related note to the two intended notes. This is particularly noticeable in the higher registers of the keyboard where the difference frequency is a note far removed towards the lower end of the keyboard and is in the easily audible range.



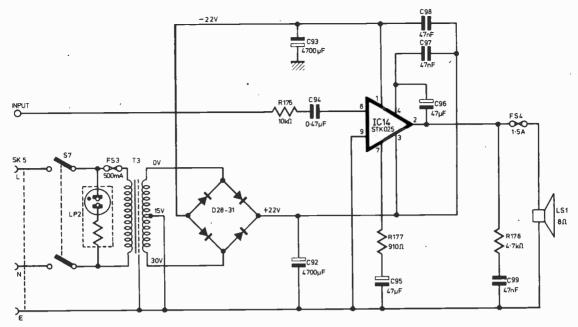


Fig. 5.4. Circuit of Amplifier

It must be admitted that the author is somewhat critical on this point although it does not affect his enjoyment of the instrument, and many observers have judged the effect to be acceptable—it can be found on the majority of commercial organs.

Intermodulation can occur throughout any amplifier chain, and is minimised by maximising the linearity of the system. The final intermodulation culprit is the output stage of the Power Amplifier, and is related to the level of crossover distortion. A number of amplifier designs provide adjustment of the output stage quiescent conditions to reduce crossover but the move to hybrid integrated modules removes this facility. Hybrid modules of course make for very easy construction and economy, and this must be weighed against the level of intermod which can be tolerated. The author's instrument continues to use such an amplifier and these notes are only included to make the situation clear, and not to deter the constructor from the use of these devices.

The circuit details of the amplifier and p.s.u. using a Sanyo STK025 are given in Fig. 5.4. The external components are mounted on a group board, and the completed unit fixed into the stool.

#### LOUDSPEAKERS

From the tone/voicing point of view the loud-speaker used with the "Joanna" is more important than the Power Amplifier. It is also important that the speaker has a relatively high power handling capacity, designed for use with electronic musical instruments where a rugged construction is vital.

If it is intended to use an existing hi-fi amplifier it is strongly recommended that the normal hi-fi speaker not be used since it is very likely that its capacity is insufficient to handle the peak power which can be and will be supplied by the amplifier when used in this application. Since the piano relies

#### COMPONENTS . . .

#### AMPLIFIER

#### Resistors

R176  $10k\Omega$ R177  $910\Omega$ 

R178 4·7Ω

#### Capacitors

C92-C93 4,700µF elect 25V (2 off)

0.47µF C94

C95-C96 47µF elect, 25V (2 off)

C97-C99 47µF (2 off)

#### Semiconductors

STK025 IC14

D28-D31 Bridge Rectifier (50V r.m.s. 2A)

#### Miscellaneous

F53 500mA fuse and panel fuseholder F54-1-5A fuse and panel fuseholder

LP2-Mains neon S7-Mains switch

SK5 3 pin miniature mains socket

SK6 mono jack socket

T3-Transformer 30V centre tapped 2A

Heatsink (R.S. 149)

Groupboard

LS1-8Ω 25W r.n.s.

#### **Touch Resistors**

R179–R214 1·5k $\Omega$  1W (36 off) R215–R239 1·8k $\Omega$  1W (25 off)

#### Music Stand

Metal Frame 36in × ¼in dia. aluminium rod Wooden crossbow 17in × ≩in × ↓in

#### Keyswitch

CPS1027 (Clef Products)

on very high peak power signals of short duration to produce an acceptable overall mean power, the ratio of peak to mean power required is much higher than in the average programme (music) played through your hi-fi system, and the danger is that you will increase playing volume without realising until you produce distortion. Distortion due to overloading the amplifier is not generally dangerous, but distortion due to overloading the speaker is fatal to that component.

The author has used 25W of amplifier (sine wave rating) and 50W (nominal) of heavy duty speaker over a period of five years with two designs of electronic piano, and whilst speakers of lower rating have been damaged in experiments, no deterioration has occurred in either a Fane POP50 or Goodmans 12P. This is not to say that other equivalent units are not available, but any speaker used should be designed for heavy duty work and its rating be treated conservatively.

#### SPEAKER VOICING

From the voicing point of view the design as presented is matched to the Goodmans 12P, mounted facing downwards in the stool. Some variation in tone quality occurs dependent on the floor covering as might be expected, but the range is acceptable. This speaker appears to apply a heavy top cut to the system, which could particularly affect the Harpsichord voice if the voice circuits were not generally bright. Experiments have been tried with smaller 8in heavy duty units, where an improved treble response is present. In order to restore the mellow piano tone, capacitors were connected across R103, R105 and R107 with values of 47nF in each case. This restored the basic tone, finely adjusted by VR3, 4 and 5, but the body of sound in the bass associated with a 12in speaker was sadly missing.

#### STOOL

Cutting details for the stool used to house the Goodmans 12P are given and the introductory photo should prove adequate for constructional guidance. The volume is below that recommended by Goodmans, and the constructor may prefer to make a more conventional cabinet to house the speaker and amplifier.

#### **FAULT FINDING**

Generally in the event of a fault occurring in one note it can be easily located to one of the 61 similar circuits, and either experienced or inexperienced substitution of components attempted. Reference to the relevant theory in the series should assist in this.

Some possibilities and causes of malfunction are quoted—it is not intended to convey the impression that such faults should be expected to occur.

Incorrect p.s.u. levels—Any general malfunction should first be followed by a check on power supply voltages. There can be a fairly widespread on voltage levels—see Part 2, but particularly a severe fall in levels can effect the operation of dividers and the master tone generator.

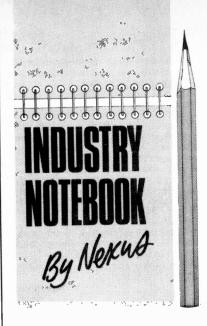
- 2. "Frog"—this could appear as a distinctive grumbling distortion of a note played on its own and may be present in all notes to some degree. Such an effect would be caused by a frequency modulation of the tone generation system at a mains frequency and indicates poor smoothing either in the p.s.u. (C1, 2, 4) or if only evident where a note is played it is likely to be a fault in either C5 or C6 on the tone generator board.
- 3. Excessive break through—If one or more notes break through predominantly into the background, output diodes D22 to D26 on the Envelope Boards should be suspected and changed for the note concerned. Some breakthrough will always be present and this is minimised by adjustment of VR3, 4 and 5, particularly the latter, on the Voice/Preamp Board. Maximum breakthrough will occur on the Harpsichord setting and is minimised by adjustment of VR2.
- 4. Spurious sustain—If a note sustains without the Sustain Pedal, the keyswitch should be checked for that note to ensure that it rests comfortably on the ground rod in the non-playing condition. Sometimes this can be confused for breakthrough.
- 5. Overload distortion. As explained earlier, this may occur in the speaker if underated or the Power Amplifier. Under exceptional circumstances it may occur in the preamplifier if the Power Amplifier is of very low sensitivity, VR9 (preset gain) should be adjusted to give zero distortion on the Piano stop for maximum level setting and heavy playing. Remember that two stops at once gives twice the power, and three stops gives three times the power. The level control on the front panel should be reduced when multi-stops are used.
- Mains hum—If this is continuous it is likely to be due to interwiring—particularly with respect to ground connections.
- 7. Intermodulation distortion—This has been covered in detail above. Do not expect perfection in this—it is an effect normally present in a wide range of electronic musical instruments.

NOTE: The mains fuse FS1 in Fig. 2.1 should be 1A. C10 and C11 should be 47nF in Fig. 2.4. It has been found that two diodes in series are necessary in the D22-D26 and D11 positions.

The transformer specified for T2 is commonly found with a 230V primary giving a high output. A resistor of 220 ohms between 0V and the junction of C1 and C2 will reduce the voltage across C1 and C2 on load.

—This concludes the Joanna series of articles.





#### SWEDISH SOLID STATE

When you come late into the field you need to find market gaps in which to become established commercially. This is never more true than in solid state technology where, to all intents and purposes, the big volume market is controlled by the giants of the industry operating internationally.

So newcomers ASÉA-HAFO, a subsidiary of the Swedish ASEA electrical group, made the decision not to compete at all in the big league but to go exclusively for the more specialised custom-built market. As ever, timing of the operation was important and the company was building up its potential just at the time that the giants were losing all interest in the custom-built market where, almost by definition, volume is low.

But even with a comparatively modest operation the start-up costs are high if an adequate service is to be offered. When I visited the ASEA Group recently I was able to see for myself where the money goes. There is a fine modern production plant, purpose built, with all services piped in below floor, clinically clean and immaculately equipped. Another building in the locality houses design staff, mask production and an opto-electronics group.

In the past three years £2 million has been invested in the operation and another £2 million is earmarked for expansion. But turnover by the end of this year is estimated in round figures at £3 million and the first investment has already been recovered.

Paradoxically, while thinking small ASEA-HAFO was also thinking big, but in different areas. With the big volume market virtually closed the products would have to

be rather special to attract profitable business in the low-volume area. The first stage in planning was therefore to set up a substantial computer-aided-design (CAD) facility geared to l.s.i. CMOS technology. The actual production process is licensed from RCA.

The most extraordinary part of the story is that the CAD department was built from scratch with no previous experience. Hardware and software could be bought but only at a price which ASEA-HAFO couldn't then afford, especially on the software side. Today there is a fine computer installation with some 200 software routines which together give a wide range of CAD programmes. Mechanisation continues into automated mask making and, after semiconductor production, into automatic testing.

Today ASEA-HAFO has a full capability in production of CMOS I.s.i., bipolar circuits, thick film hybrids and opto-electronics. Typical custom orders are for 10,000 units in CMOS and the business is going well with a design capacity of one circuit per

month.

ASEA-HAFO could not have succeeded without the massive backing of the ASEA Group, founded in 1883 and dubbed the GEC of Sweden. With 40,000 employees it ranks 11th in the world league of electrical companies. From a mass of company statistics I think the most significant is that sales per person employed in 1974 was £14,000, a fine example of productivity and heavy investment in plant and machinery.

One of the most fascinating products I saw was the ASEA Industrial Robot which, suitably programmed, can do repetitive jobs like pallet loading of components and some of the hardier tasks in bad environments such as welding and die-casting. A wonderful production aid which can work 24 hours a day without fatigue and clearly is designed to displace people. When I asked whether there was trade union pressure against its introduction I was told the unions welcomed it because it saves the members doing dull and dirty work, an attitude which has helped Sweden become prosperous.

#### MORE THAN A HANDSHAKE

If all has gone well, the dramatic handshake in space between American astronauts and Russian cosmonauts will have taken place (it was due on July 15) following the docking of the Apollo and Soyuz spacecrafts.

The ceremonial aspects of the occasion will perhaps have added credence to current political moves at detente between the two nations. But after this phase of the operation, which you may have seen on television, the real work began. In the American case the astronauts were due to conduct some 20 scientific investigations over the following five days. And of several groups of experiments the most important for the electronics industry was that involving the processing of semiconductor materials in a high temperature furnace under conditions weightlessness.

This series of experiments was started during the *Skylab* missions in 1973 and 1974. Results were promising. An improved furnace was carried on the *Apollo-Soyuz* mission as part of the programme to confirm earlier *Skylab* findings If crystal imperfections can be eliminated by "growina" them under weightless conditions it should be possible to produce more transistors per given area of material and thus produce integrated circuits with the ultimate in component density.

The orbiting factory may not be too far in the future. It is quite conceivable that within a few years we shall not think it at all unusual that deliveries of semiconductor crystals should arrive regularly on earth from outer

#### MARITIME COMSATS

A more immediate prospect is the realisation of commercial ship's communication by marine satellite. W. Martin Lovell, deputy technical director, Marconi Space and Defence Systems, gave an up-date on the situation of the European MAROTS programme at a Communication Conference in Rotterdam.

Starting as a British national programme, MAROTS has now been transferred to the European Space Agency (ESRO). Structure thermal tests are due for completion this year, the engineering model completed by mid-76, the qualification model completed early in 1977 and the flight model ready for launch by *Delta 3914* in October 1977.

As well as being responsible for the communications module, Marconi are pressing ahead with a shipborne terminal for general marine use. The big market will not come until the 1980's when the system has been demonstrated and successful and ship owners decide they can't afford to be without it.

Nobody is better qualified in marine communication than Marconi, the first company in the world to exploit maritime communication by radio 75 years ago.



input impedance, so as not to affect the audio and stereo switching signals and (b) it responds to negative signals, yet does not require an additional negative rail.

One would find that due to the large spread of parameters, one would have to alter slightly the values of R1, which could be regarded as the gain control. R3 may also have to be changed, depending on the meter employed.

To adjust the circuit, it is required to tune into the far right of the f.m. scale, far from any broadcast station, and adjust the preset until the meter indicates

centre.

Since no modification to the meter itself is required, this circuit lends itself easily for switching from centre tuning function to peak signal tuning function.

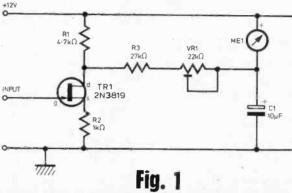
> M. Greenfeld, Leeds.

# F.M. CENTRE TUNING METER N v.h.f./f.m. broadcasts, the bes

In v.h.f./f.m. broadcasts, the best reception is arrived at when one tunes to the centre of the S curve of the detector stages, i.e. 0V d.c. output from the detector coil (when conventional detector is employed).

Fig. 1 shows a circuit which allows conversion of a signal strength meter into a centre tuning indicator; the only modification required is the addition of a centre mark on the meter scale.

The circuit gets its input directly off the detector output (before any coupling capacitors). The f.e.t. serves two purposes: (a) very high



## TWO FUNCTION R/C

THE circuits shown in Figs. 1 and 2 were developed as a result of the 1971 series "Logical Radio Control". I had difficulty in getting the coder described in the series to work reliably and so it was modified so that the train of monostable-generated pulses was triggered by an astable multivibrator.

This gave a system of fixed cycle time which, if the individual pulse lengths were kept short, made the generation of a synchronising pulse unnecessary as long as a suitable decoder was used.

Difficulty in obtaining the retriggerable monostable (F9601) resulted in the development of a 2-channel decoder.

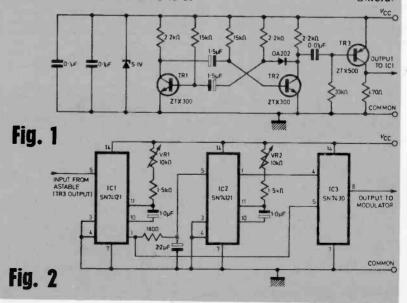
The coder can easily be built on Veroboard or a p.c.b. can be made up. Due to the short rise and fall times of the pulses, short wiring is desirable. Decoupling of the supply with  $0 \cdot 1 \mu F$  capacitors is also advisible. The prototype was powered from a 6V battery with an OA81 diode connected in the +ve lead.

Testing with an oscilloscope should show a waveform of 36ms pulses (at the collector of TR3) 0.3ms apart. The output of IC3 should be pulses whose length will depend on the setting of VR1 and VR2. The length of these pulses should be variable from approximately lms to 7.5ms.

For those requiring control over more than two functions it would be

possible to add at least one more if not two more monostables but care must be taken that the maximum total pulse length does not exceed the fixed cycle time of 36ms.

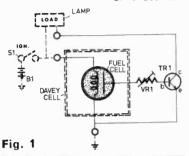
S. D. G. Tucker, Oxford.



# PATENTS REVIEW...

#### CAR GAS DETECTOR

BP 1 387 488



Claims for a simple gas detector primarily for use in motor vehicles is outlined in BP 1 387 488 filed by TRW Inc.

Exposure to anything over fifty parts per million of carbon monoxide for a period in excess of one hour can affect the capability of a driver, even though twice this concentration at three hours' exposure will be needed to cause perceptible effects. Thus, a leaky exhaust pipe can take the edge off a driver's performance without him realising it.

The circuit Fig. 1 shows a fuel cell which includes an electrode in the form of a heating element. This element is connected between supply, via ignition switch, and car earth to maintain a temperature of about 600°C. The element is surrounded by a porous refractory shroud of silica and alumina, the shroud being coated by a porous layer of tin oxide or other catalytic metal oxide.

A second electrode is connected via VR1 to the base of TR1. The latter has its collector-emitter circuit connected to the warning lamp and earth respectively.

The catalytic coating causes oxidation of any carbon monoxide present and the release of ions. This causes current to flow across the electrodes and TR1 to conduct and operate the alarm.

The value of VR1 is adjusted to control triggering of the alarm, for instance at a concentration of carbon monoxide in the atmosphere of 100 parts per million.

#### LOUDSPEAKER FILTERING

In BP 1 385 346, Saba Schwarz-walder Apparate-Bau-Anstalt August Schwer Sohne GmbH refers to the undesirable group radiation or interference effects which occur when several identical loud-speakers are used in a single small cabinet.

Where it is required to use one high frequency tweeter or midrange unit with more than one identical bass speakers, it is unacceptable to have the bass speakers sharing the same load at the region of crossover with the middle or high range units. Although the bass units may share the same load over a part of their operating frequency range, all but one of them must be rolled off to zero well before the crossover range. According to the inventors, this roll-off should be at around 12dB per octave and start at least an octave below the crossover

A suitable circuit is shown in Fig. 1. Two low frequency speakers of identical type are independently filtered. Loudspeaker LS1 is in series with inductance L1 and in parallel with C1. LS2 is in series with L2 and in parallel with C2.

As shown in Fig. 2, the filter for LS1 is a low-pass filter with a top cut-off of substantially  $f_0$ . The filter has a slope of 12dB per octave and overlaps the curve of the high pass filter for a tweeter. The filter for LS2 also acts as a low pass filter but with a cut-off at  $f_1$  which lies more than an octave lower than  $f_0$ . Thus, only LS1 is effective at the

BP 1 385 346

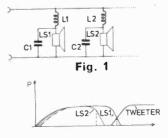


Fig. 2

overlap region and there is no group effect.

TV SOUND UNIT BP 1 389 228

In BP 1 389 228, Dinosaur Elec-Limited,\* assignees of tronics Anthony Burton of Windsor, claim commercial protection for a TV sound unit. This enables high fidelity sound to be derived from a conventional TV receiver by picking up stray i.f. sound radiation, demodulating it and feeding it to the auxiliary socket of a conventional hi-fi system. The patent appears to cover the commercially available Telefi, and the claims are probably broad enough to monopolise most other stray pick-up systems, including last month's PE design.

As previously explained in this column, British Patent Law is not such an ass as to prevent amateur constructors from conducting private experiments in their own homes (for instance based on circuitry published in the electronics press); but any home constructor who feels inclined to move on into even low-scale production of a patented device should take legal advice before doing so.

The Dinosaur patent explains how the tuner output of a conventional television receiver is passed to an i.f. amplifier and vision detector. This feeds a vision signal to the video circuits and a sound carrier signal to a sound i.f. amplifier, a limiter and a frequency discriminator.

The sound carrier signal is equal to the difference between the vision and the sound carrier frequencies and is frequency modulated with sound information. Inevitably there is stray radiation from the i.f. strip, and in 1971 the inventor lodged a patent claim for the straightforward but clever approach of capturing some of this stray signal by a tuned pick-up and amplifying and demodulating it to provide a high quality audio signal.

The patent covers all types of pick-up, including a coil or aerial and various transformer circuity for connecting the pick-up to the external amplifier and demodulator.

\*The above mentioned patent is now held by Rola Celestion Ltd., manufacturers of the "Telefi" sound unit.

Copies of Patents can be obtained from the Patent Office Sales, St. Mary Cray, Orpington, Kent. Price 33p each

# TRANSDUCERS - Light

PART 6 und Opto electronic Devicest Light Speed Force Load Sound Frequency Distance Heat

This final article of the series looks at some opto electronic devices and digital encoding transducers for monitoring shaft rotation or position.

#### **SOME BASICS OF LIGHT**

Technically, the term "light" means electromagnetic radiation having a frequency or wavelength which can be perceived by the human eye. The percentage of the available electromagnetic spectrum occupied by visible radiation is in fact very small and covers the wavelength range from 4,000 to 7,000 Å (Angstrom units). The wavelength and frequency of the electromagnetic waves are related to the velocity of light by the well-known relationship:

Velocity = Frequency \ Wavelength

The velocity of light is approximately  $3\times10^8$  metres/second and is a constant. Consequently short wavelengths correspond to the high frequencies and vice versa. The wavelength range above, quoted in Å can be related to more common units of length by noting that

$$1 \text{ Å} = 10^{-8} \text{cm} \text{ and } 10,000 \text{ Å} = 10^{-3} \text{mm} = 1 \mu \text{m}$$

Thus the wavelength range of visible light is from 0.4 to 0.7 thousandths of a mm which is very short when compared to the radio wavelengths in everyday

There are two related aspects of light that are used, together or separately, to explain many phenomena associated with light. Optical phenomena such as interference and diffraction patterns are explained by a wave theory such as that involved with radio waves. However, other phenomena are better explained by the quantum theory in which particle-like aspects of light are considered. According to this theory, light can be considered as discrete quanta or packets called photons.

These photons are uncharged particles each having an energy which depends only on their frequency or wavelength as given by

Energy (in electron volts) = 
$$\frac{h c}{\lambda}$$

where h is a constant (known as Planck's constant), c is the velocity of light and  $\lambda$  is the wavelength of the light.

Notice that this relationship shows the energy as increasing as the wavelength decreases. The energy associated with any particular wavelength of light can be obtained from the above equation, which can be implified by putting in the values of c and h thus,

Energy 
$$E$$
 (in eV) =  $\frac{12,400}{\lambda \text{ (Angstrom)}}$ 

Table 6.1. Optical energy gaps of photodetector materials

Material	Chemical Symbol	Energy Gap at 300°K (eV)
Cadmium Sulphide	CdS	2.4
Cadmium Selenide	CdSe	1.7
Gallium Arsenide	GaAs	1.4
Silicon	Si	1.1
Germanium	Ge	0.7
Indium Arsenide	InAs	0.43
Lead Sulphide	PbS	0.37
Lead Selenide	PbSe	0.26
Indium Antimonide	InSb	0.23

For the visible spectrum limits this will give us Violet light ( $\lambda \sim 4.000 \text{ Å}$ ). F = 3.1 eV

Violet light ( $\lambda \approx 4,000 \text{ Å}$ ), E = 3·1 eV Red light ( $\lambda \approx 7,000 \text{ Å}$ ), E = 1·77 eV

Incidentally, the electron-volt unit of energy is the energy acquired by an electron when raised through a potential difference of one volt. Compared to the more familiar units of energy, the Joule, the eV unit is very small and the two units are related by

1 Joule = 
$$6.24 \times 10^{18} \text{ eV}$$

Photons can be emitted from the atoms of certain materials when electrons within change their energy from a high level to a lower level. The energy of the emitted photon will be equal to the energy lost in falling from the high to the low level. This energy difference is known as the energy gap and is also usually expressed in eV. Photons can also be absorbed by a variety of materials. If the energy of the entering photon is greater than the inherent energy gap of the material the valence electrons in the material can be excited and lifted up to the conduction band level with the resultant creation of free electron (and hole) current carriers. This particular process is in fact the basis of operation of all photodetectors, regardless of the particular media. Typical energy gaps for several materials used in the manufacture of photodetectors are shown in Table 6.1.

#### **BASIC PHOTO-DETECTOR PARAMETERS**

The photodetector transducer is the heart of most opto-electronic systems and the selection of the best detector for the particular application is often very important. Obviously the photodetector must be considered in relation to the light source, which might be

<sup>\*</sup> North Staffordshire Polytechnic

the sun, a filament lamp, a light emitting diode or a laser. Photodetectors can be divided into two classes.

- 1. Thermal Detectors in which the radiation is absorbed, converted into heat, and the detector responds to the change of temperature.
- 2. Quantum Detectors which respond directly to the incident photons. This class can be subdivided into
  - (a) Photo-emissive, where the incident photons release electrons from the surface of the detector material. This phenomena occurs in vacuum, as in vacuum photo-diodes or phototube multipliers.
  - (b) Photo-voltaic, where a voltage is self-generated when light strikes the detector without any external bias. The solar cell is a well known example of this type but photo-diode operation is also based on this effect.
  - (c) Photo-conductive, where the conductivity of the photo sensor changes as a function of the incident light. These can be undoped devices such as photoresistors, or doped photoconductors such as photodiodes and similar devices.

In the selection of a photodetector transducer it is usually necessary to consider the following four parameters in relation to the application.

#### SPECTRAL RESPONSE

This is really a measure of the wavelength range over which the detector can respond. If the lightsource spectrum ranges from, say, 4,000 to 8,000 Å with a peak at 6,000, then a detector with a similar response, or at least an appreciable spectral overlap, should be used.

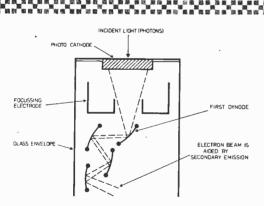


Fig. 6.1. Operating principle of the photomultiplier tube in diagrammatic form

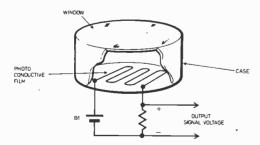


Fig. 6.2. Simplified construction and circuitry associated with a CdS cell

#### **FREQUENCY RESPONSE**

This relates to the speed with which the detector can respond to changes in the level of the incident light and is important when modulated radiation is to be used.

#### NOISE

Any random fluctuation of output current or voltage constitutes an unwanted noise signal. The term dark current is often used to denote the d.c. photodetector output current that exists under dark conditions whereas the noise current is the random a.c. output. Noise is directly proportional to the square root of the detector area.

#### RESPONSIVITY

This refers to a detector's sensitivity or output per unit light input. The responsivity can be expressed in various ways such as output current in milliamperes per input radiant light flux in watts.

#### SIGNAL/NOISE PERFORMANCE

The minimum detectable signal level will depend on the amount of noise produced within the photodetector. The performance is sometimes specified in terms of a Noise Equivalent Power (NEP) which is defined as the input (signal) power required to produce a signal/noise ratio of unity when using a bandwidth of 1Hz. Since the noise depends on the area of the photodetector and the bandwidth being considered, care must be taken when comparing different devices.

A variety of common photodetectors will now be dealt with in some greater depth.

#### **PHOTOMULTIPLIERS**

One of the traditional devices for detecting light is the photomultiplier tube. These devices provide high sensitivity and speed and the operating principle is illustrated in Fig. 6.1. Incident light releases electrons from a photocathode which has peak spectral response in the 4,000 to 5,000 Å range. The electrons are focused and accelerated towards the first dynode by appropriate electric fields and here additional electrons are released by the process of secondary emission. The number of secondary electrons emitted per incident primary electron, ranges between 1 and 5 typically and by using a series of dynodes, current gains (overall) of 10<sup>5</sup> or more can be obtained.

Photomultipliers and their newer counterparts, continuous channel multipliers, are usually the best choice for applications requiring extremely low-level light detection with frequency responses greater than a few megahertz. However, a wide range of solid state photo sensor/amplifier configurations are possible which can give comparable speed or sensitivity, and superior performance can often be achieved in low speed applications.

#### **BULK PHOTOCONDUCTORS**

In these devices a thin film of photo-conductive material is exposed to the incident light. Provided the energy is sufficient the photons can release electron-hole pairs in the material and the resistance of the cell therefore falls as the intensity of the light increases. An external battery and circuit is required for these devices and Fig. 6.2 illustrates the principles outlined for one particular device, the CdS cell.

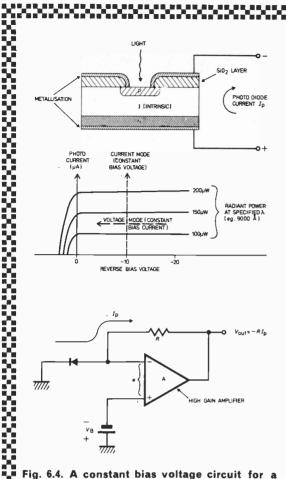


Fig. 6.4. A constant bias voltage circuit for a photodiode

Such cells exhibit an approximately exponential variation of resistance with ratios of dark to light resistance of about 1,000:1. The response is relatively slow, especially when the light source decreases or goes off, since it takes longer for the electrons to "drain" back to the valence band due to trapping in imperfections within the material. Resistive photo conductors also exhibit temperature effects which vary with light level and generally speaking the faster response materials are less temperature stable than the slow response types.

#### **PHOTODIODES**

In this case the detector uses a doped semiconductor p-n junction and the incident radiation generates hole/electron pairs providing the photon energy exceeds the energy gap requirements. The pair generation will occur at various depths depending on the energy of the photons and the nature and thickness of the materials. If the pairs are produced outside the junction depletion region there is a greater probability of them recombining so the depletion region is often widened by increasing the reverse bias across the diode.

This also decreases the transit time of the holes and stectrons. Sometimes an intrinsic layer is interposed between the p and n regions and this gives the device a p.i.n. structure which also produces a wider depletion region. The p.i.n. photodiode has a lower junction capacitance by virtue of the increase in the "plate" separation and this makes the device much faster than the conventional p-n type.

The p.i.n. structure also exhibits low noise and dark current plus greater efficiency at the longer wavelengths.

A typical structure is shown in Fig. 6.3 together with typical current-voltage characteristics. Operation with fixed reverse bias gives a current mode (high internal resistance) characteristic in which the output current is a linear function of incident radiation over about ten decades. Operation at constant bias current is also possible but gives more distortion in "linear" applications and a greater dependance on environment and bias current stability.

Fig. 6.4 illustrates the current mode where the reverse bias voltage across the diode is provided by the amplifier action and supply  $V_B$ . The virtual earth principle resulting from  $A \geqslant 1$  means that e is almost zero. The diode bias is thus fixed at  $V_B$  volts and the photodiode current,  $I_B$ , is used to drive the amplifier as a current to voltage converter, to give an output voltage proportional to diode current.

#### **AVALANCHE PHOTODIODE**

The avalanche photodiode is a special photodiode designed for operation in the avalanche breakdown region in order to achieve internal photo-current multiplication. The electrons, produced by pair generation, drift into the avalanche region of the device where they are rapidly accelerated by an intense electric field. The high velocities give a high probability of generation of additional electrons by impact ionisation. The ionisation rate is a strong function of applied voltage because the electric field strength determines the energy of the generated current carriers and hence the overall current multiplication. The avalanche photodiode is a sort of solid state equivalent to the photomultiplier tube.

#### **SOLAR CELLS**

The advent of the space age resulted in extensive research into the area of direct conversion of solar energy into electrical energy. More recently, a growing need for energy conservation has come about and it seems likely that solar cell development will continue for certain applications. To be a good energy converter the solar cell spectral response should match the sun's spectrum which peaks at about 5,000 Å and yields an irradience level of about 100mW/cm² at the earth surface. For energy conversion applications it is also desirable to have a high power conversion efficiency.

Typical solar cells for such applications must have low internal resistances and the basic structure, operation and fabrication are similar to the silicon photodiode. The doping levels and areas are much higher however and the top layer of the junction is very thin to extend the U.V. response to match the solar spectrum. In some cells the contact resistance is decreased by using a fine mesh contact and normally the cells are operated without bias in the photo voltaic mode.

Solar cells are often used in photographic exposure meters. In this application the solar cell is superior to a photodiode due to the larger area and therefore high

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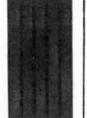
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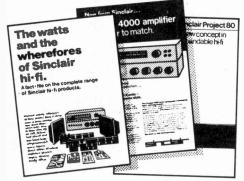
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1N21			1.15	BY213	0.25			Z8170	0.10
1N23 1N85	0-85 0-88	AFZ12 ASY26	2-00 0-25	BYZ10 BYZ11	0-45			Z8271 ZT21	0·18 0·25
1N253	0.50		0.38	BYZ11	0-40 0-40			ZT43	0.25
1N256	0.50		0-25	BYZ13	0.42			ZTX107	0.12
1N645	0-16	ASY29	0.30	BYZ15	1.25				0-08 0-13
1N725		ASY36	0-25	BYZ16	0-60				0.13
1N914		A8 Y 50	0.20	BZY88	0.10	0AZ222	0-45	ZTX500	0.13
1N4007		A8Y51	0.40	C111	0.55	OAZ223	0.45	ZT X 503	0.18
18113	0.25	ASY53	0.20	CR81/05	0-35	OAZ224		ZTX531	0.25
18202	0.28	ASY55 ASY62	0.20	CRS1/40 CS4B	0.50 1.90	OAZ241	0-25	INTEGR	ATED
264371	0.40	A8 Y 66	0.38	C810B	3-50	OAZ242 OAZ244	0·15 0·25	CIRCUIT	
2G381	0.22	A8Z21	1.00	DD000	0.15	OAZ246	0.15	7400	0-16
20414	0.80	ASZ23	0-75	DD003	0.15	OAZ290	0.88	7401	0.16
20417	0.25	AU104 AUY10	1.00	DD006	0.25	OC16	1.00	7402	0.16
2N404	0.22	BC107	0.14	DD007 DD008	0.40 0.38	OC16T	1.00	7403 7404	0-16 0-26
2N697	0.16	BC108	0.18	GD3	0.38	OC19 OC22	0.50	7405	0.22
2N698 2N706	0·80 0·12	BC109	0.14	GD4	0.10	OC23	1.25	7406	0.42
2N706A	0.12	BC113 BC115	0-15 0-20	GD5	0.83	OC24	1.10	7407	0.42
2N708	0.15	BC116	0.20	GD8 GD12	0.25 0.10	OC25	0-40	7408 7409	0.28 0.28
2N709	0-40	BC116 BC116A	0.28	GET102	0.50	OC26 OC28	0-40 0-66	7410	0.16
2N1091 2N1131	0.55 0.25	BC118	0.20	GET103	0-40	OC29	0.65	7411	0.25
2N1132		BC121 BC122	0.20	GET113		OC30	0-40	7412	0.80
2N1302	0.18	BC122	0.68	GET114	0.80	OC35	0.55	7413	0.86
2N1303	0.18	BC126	0-65	GET115 GET116	0.90 0.85	OC36	0.60	7416 7417	0-36 0-86
2N1304 2N1305	0.28 0.22	BC140	0.55	GET120	0.50	0C41 0C42	0-85 0-40	7420	0-16
2N1306	0.28	BC147	0-10	GET872	0-80	OC43	0-70	7422	0.25
2N1307	0.28	BC148 BC149	0.08 0.10	GET875	0-40	0.044	0.20	7423 7425	0.87
2N 1308		BC157	0-14	GET880 GET881	0.00	OC44M OC45	0·17 0·20	7425	0.87 0.87
2N2147 2N2148	0.78 0.60	BC158	0.12	GET882	0.35	OC45M	0.20	7428	0-40
2N2146 2N2160	0.78	BC169	0.68	GET885	0.40	OC45M OC46	0.27	7430	0-16
2N2218	0.28	BCV31	0·14 0·45	GEX44 GEX45/	0.08	OC57	0.60	7432	0-87
2N2219		BCY32 BCY33 BCY34	0.85	GEX941	0.45	OC58 OC59	0.60	7433 7437	0-87 0-87
2N2369 2N2444	A 0.16 1.99	BCY33	0.88	GJ3M	0.50	OC66	0-60	7438	0-87
2N2613	0.75	BCY34	0.45	GJ4M	0.50	OC70	0-18	7440 '	0.22
2N2646	0.50	BCY38 BCY39	0.55 1.50	GJ5M	0.25	OC71	0.18	7441AN	0.92
2N2904	0-20	I RCV40	0.80	GJ7M HG1005	0.50 0.50	0072	0.28	7442	0-79 0-16
2N 2904		BCY42	0.30	H8100A	0.20	OC73 OC74	0-50 0-30	7451	0-16
2N2906 2N2907	0.23	BCY42 BCY70 BCY71	0.18	MAT100	0.20	OC75	0-80	7453	0.16
2N2924	0.13	BCY71	0.22	MAT101	0.25	OC76	0.80	7454 7460	0.16
2N2925	0-15	BCZ11	0-65	MAT120 MAT121	0-25	OC77 OC78	0-54 0-25	7470	0-16 0-86
2N2926 2N3054	0·12 0·48	BD121	1.00	MJE340	0-47	OC79	0.20	7472	0.88
2N3055	0.45	BD123 BD124	1.00	MJE520	0.63	OC81	0.29	7473	0.41
2N3702	0.11	BDY11	0.65 1.45	MJE2955 MJE3055	1.27	OC81D	0.28	7474 7475	0.42
2N3705	0-15	BF115	0.20	MPF102	0.40	OC81M OC81DM	0.20	7476	0-45
2N3706 2N3707	0-11 0-18	BF167	0.25	MPF103	0.86	OC81Z	0-45	7480	0.60
2N3709	0.10	BF173 BF181	0-28 0-85	MPF104	0-35	OC82	0.28	7482	0-87
2N3710	0.11	BF 184	0.22	MPF105 NKT128	0-36 0-45	OC82D	0-25	7483 7484	1·10 1·00
2N3711 2N3819	0-11 0-88	BF185	0.22	NKT129	0-80	OC83 OC84	0-27 0-30	7486	0-47
2N4289	0-80	BF194 BF195	-0-10	NKT211	0.25	OC114	0.88	7490	0.55
2N5027	0.53	BF196	0·18 0·15	NKT213	0.25	OC122	1.00	7491AN 7492	1.00
2N 5088 28301	0-88 0-59	BF197	0.15	NKT214 NKT216	0.24	OC123 OC139	1·10 0·40	7493	0·70 0·70
28301	1.15	BF861	0.25	NKT216 NKT217	0.45	OC140	1.14	7494	0.80
28501	0-75	BF898 BFX12	0.25	NKT218	0.45	OC141	0.80	7495	0.80
28703	1-00	BFX13	0.26	NKT219 NKT222	0.88 0.80	OC169	0-20	7496 7497	0-95 3-87
AA129 AAZ12	0-20 0-75	BFX29	0.28	NKT224	0.25	OC170 OC171	0.80	74100	1.89
AAZ13	0-12	BFX30	0.28	NKT251	0-24	OC200	0-54	74107	0-45
AC107	0.51	BFX35 BFX63	0-98 0-50	NKT271	0.20	OC201	1.00	74110	0-58
AC126	0.25	BFX84	0.25	NKT272 NKT273	0.20	OC202 OC203	0.90	74111 , 74118	0.86
AC127 AC128	0.25 0.15	BFX85	0.28	NKT274	0.20	OC204	0-55 0-65	74119	1.68
AC187	0.21	BFX86	0-25	NKT275	0.25	OC205	1.00	74121	0.50
AC188	0-20	BFX87 BFX88	0·25 0·24	NKT277	0.20	OC206	1.10	74122	0.70
ACY17	0-40	BFY10	0.50	NKT278 NKT301	0-25 0-85	OC207 OC460	1.00	74123 74141	1.00 0.90
ACY18 ACY19	0-27 0-27	REVII	0.50	NKT304	0.75	OC450 OC470	0.20	74145	1.26
ACY20	0.22	BFY17 BFY18	0-40	NKT403	0.70	OCP71	1.20	74150	1.75
ACY21	0.22	BFY19	0-45	NKT404 NKT678	0.66	ORP12	0.60	74151 74154	1-00 2-00
ACY22 ACY27	0·16 0·25	BFY19 BFY24	0-45	NKT713	0-80	ORP60 ORP61	0-55 0-48	74154	1.00
ACY28	0.25	BFY44	1.00	NKT773	0.25	8X68	0.20	74156	1.00
ACY39	0.78	BFY50 BFY51	0.21	NKT777	0-88	SX 631	0.45	74157	0.95
ACY40	0.22	BFY52	0.20	OA5	0.72	BX 635	0.55	74170 74174	2.52
ACY41 ACY44	0·22 0·32	BFY53	0.17	OA6 OA47	0.12	8X640	0.75	74175	1.10
AD140	0.50	BFY64 BFY90	0.86	0A70	0.10	8X641 8X642	0-75	74176	1.26
AD149	0.50	BSX27	0.81	OA71	0.20	8X644	0-85	74190 74191	2·00 2·00
AD161 AD162	0-44	BSX 60	0-98	OA73 OA74	0-15 0-15	8X645		74192	2-00
AF106	0.80	BSX76	0-18		0.10	TIC44	0.29	74193	2.00
AF114	0.25	BSY26	0.17	OA81	0.18	V15/30P	0.75	74194	1.80
AF115 AF116	0.25	BSY27 BSY51			0.15	V30/201P		74195 74196 74197	1·10 1·20
AF116	0-25 0-24	BSY95A	0.18		0.15	V60/201		14101	1.20
AF118	0.57	BSY95	0.12		0.07	V60/201P		74198	2.77
AF119	0.20	BT102/50		OA91	0-07	XA101	0.20	74199	2.52
AF124 AF125	0-80	BTY42	0.75	OA95 OA200	0-07	XA102 YA151	0-18 0-15	Plug in soc	kets
AF126		BTY79/1	00 R	OA200 OA202	0-08	XA151 XA152	0.15	-low prof	lle
AF127	0-80		0-75	OA210	0-20	XA161	0.25	14 pin D]]	0-15
AF139 AF178	0-41	BT Y79/4	00 R	OA211	0-35	XA162	0.25	16 pin DII	
AF178 AF179	0-85	BY100		OAZ200 OAZ201	0-50	XB101 XB102	0-43		0-17
AF180	0-55	BV126	0-14	OAZ202	0.45	XB103	0-35		
AF181		BY127	0.12	OAZ203	0.45	XB113	0.30		
AF186	THE RESERVE	BY182	STREET, SQUARE,	NAME AND ADDRESS OF THE OWNER, WHEN PERSON OF THE OWNER, WHEN PERSON OF THE OWNER, WHEN PERSON OF THE OWNER,	-	XB121	0-48		
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#### **PHOTOTRANSISTOR**

This device combines the ability to detect light and provide gain in a single device. The structure is similar to that of a conventional planar transistor except that a window or lens is provided in the case. The incident light generates hole-electron pairs in the vicinity of the relatively large reverse biased collector-base junction and these give rise to additional current by virtue of transistor action.

The dark current is also magnified by this same mechanism so that there is no basic improvement in signal/dark current ratio. The collector-base capacitance also affects the frequency response which is usually inferior to that obtained from a photodiode/amplifier combination. Variation of transistor gain  $(\beta)$  with current also affects the linearity and consequently their best area of application is in "ON-OFF" situations such as punched card and paper tape readers.

Linear arrays of devices have been produced for such applications. More recently photo field effect and related devices have been developed for specialised uses and light activated thyristors are yet another example of the growth that is taking place. Infra red detectors have also been developed, largely due to military applications in the first instance and many of these require operation at low temperatures.

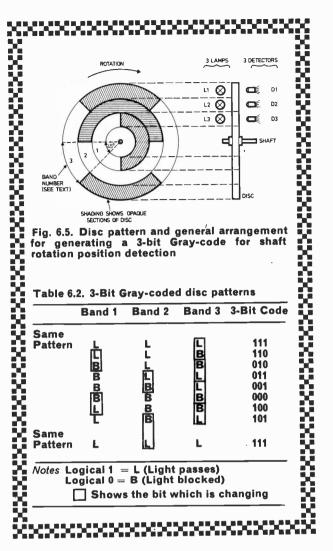
#### LIGHT SOURCES

The most common form of light source is the incandescent lamp which is available in a wide variety of shapes and sizes. Gas discharge lamps are also well known and widely used in neon displays and fluorescent lighting systems, but there is a growing area of applications for low cost, solid state light sources such as the light emitting diode (l.e.d.).

A multitude of crystals have been used to produce l.e.d.s and today various coloured light sources and displays are available. Combined light source-detector combinations are also available and are usually known as opto-couplers. These can be based on any of the available opto devices but the l.e.d. photodiode (or phototransistor) is perhaps the most common. These devices can give a high degree of electrical isolation and find application in such areas as patient monitoring and diagnostic medical equipment interfaces. Combinations of devices have also been developed as photopotentiometers and photo-choppers but the photon-coupling is a common feature of all these devices.

#### **DIGITAL ENCODERS**

A fairly common requirement is to monitor the position of a rotating shaft and one way of doing this is to use a shaft encoding disc which carries a particular optical pattern. The pattern is arranged to interrupt a series of light signals in such a way that a digital code is generated that represents the shaft position. The principle is illustrated in Fig. 6.5 which shows a pattern for generating a 3-bit Gray code. This code is so arranged that only one bit of the code changes at any one time and thus the alignment problems of binary-coded discs are avoided. The three circular bands on the disc are broken



into segments and it is assumed that the shaded segments cut off the light sources from their respective photo-detectors when they fall between the lamps and detectors (such as Ll, Dl). Starting with all bands "transparent", as in the region where the band numbers are shown, and assuming clockwise rotation of the disc with the letter L corresponding to the passage of "light" and B for "black" (or dark-current conditions in the detector) we get the sequence of events shown in Table 6.2.

The position of the disc is identifiable, to within any one sector of 45°, by a unique 3-bit code. As the disc rotates only one bit change occurs at the sector boundaries due to the relative position of the segments within the three bands.

Shaft speed can be determined by using a series of holes, equally spaced around the periphery of a disc, to chop a light beam and so produce a pulse train whose repetition rate is proportional to speed of rotation. Obviously N holes in a disc rotating at n revs. per second will produce an output having a frequency of  $N \times n$  pulses/sec.

Light sensitive devices can obviously be used in a wide variety of applications. For example it is possible by suitable diffraction gratings or discs, to produce interference patterns involving alternate light and dark

bands. These patterns of "lines" can be detected and/or counted by opto-devices and using these techniques it is possible to measure angles of inclination of the gratings, or rotation angles of discs and shafts, to very high accuracy.

#### **ACKNOWLEDGEMENTS**

In a short introductory series it is impossible to cover all the transducers and related devices that are in current use. However, the series should provide the reader with some basic information which can be extended by selected reading. The author has been influenced by various people, papers and books over a number of years and several manufacturers have been very helpful in supplying details of their devices. The author wishes to acknowledge all these sources and suggests the following references for study by those who wish to pursue the subject further:

Practical Instrumentation Transducers by F. J. Oliver (PITMAN)

Transducer Measurements by Arthur (TEKTRONIX)

Piezo Electric Ceramics (Phillips, Application Book) Piezo Electric Air Transducers (Mullard App. Notes

Industrial Linear and Non Linear Resistors (TP1174 Mullard)

Basic Electronic Instrument H/B by C. F. Coombs (McGRAW HILL)

Strain Gauges, Theory and Handling by H. Kiihl (PHILLIPS)

Understanding Thermocouples: Instrument and Control Engineering (October, November 1968)

I.E.E. Conference on Servocomponents 1969

Servo Systems. Electronic Data Library Vol. 2 (Morgan-Grampian 1969)

The E-Cell Application No. 5, Electron, 1st June 1972 The Question of Thermistors, Electron, 13th July 1972 Thermistor Manual — Fenwal Electroncis

Instrument Technology by E. B. Jones (NEWNES-BUTTERWORTH).

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50mA	£4.50
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5m A	. £3.0			
10mA .	€3.0			
20mA	€3.0			
50m A	€3.0			
100mA	€3.0			
150mA	£3.0			
200mA	. £3.0			
300mA				
500mA				
	. £3 0			
	. €3.0			
ZA DC	£3.0			

10A DC 3V DC



#### CLEAR PLASTIC MODEL MR 85P

JEAN PLAS Size: 120 x 110n	
50u A	€6.00
100uA	€5.95
200uA	€6.90
500uA	€5.80
50-0-50uA	€5.95
100-0-100uA	£5.90
500-0-500uA	
1mA	€5.75
1-0-1mA	€5.75
5mA	€5.75

	€6.00	F 5.7.5 W.	
	£5.95	1000000	
	€6.90	¥ VU	22
500uA	€5.80	- N	
50-0-50uA	€5.95	THE NAME OF THE PARTY OF THE PA	
100-0-100uA	£5.90		
500-0-500uA	€6.55		
1mA	£5.75	HERRICAL TO STORY	10.00
1-0-1mA	€5.75	50V DC	£5.75
5mA	€5.75	150V DC	
10mA	€5.75		
50mA	£5.75	15V AC	
100mA	£5.75	300 V A C	
500mA	€5 75	5 Meter 1mA.	
1A DC	€5 75		£7.10
5A DC	£5.75	1AAC	£5.75 *
15A DC	€5.75	SAAC	€6.75 *
30A DC	£6.95	10A AC	€5.75 *
18V DC	€5.75	20A AC	£5.75 *
20V DC	€5.75	30A AC	£5.75 °

#### MODEL PE70

EDGWISE MU	
Size: 90 x 34mr	n
50u A.	€4.55
100uA .	€4 50
200uA .	€4.45
500uA	£4.30
50-0-50uA	£4.50
100-0-100uA	£4.45
1mA	€4.25
10V DC	£4.25
300 V D C	£4.35
VU Meter	£5.50



POSTAGE / PACKING & INSURANCE 15p

#### BAKELITE MODEL S80 Enlarged Window

MUEF11F 11	HODEL OF	,0 5
ize: 80 x 80m		
50uA	€4.95	
100uA	€4.90	
500uA	€4.60	100
50-0-50uA	£4.80	
100-0-100uA .	. £4.85	
1mA	€4.60	0 ,
1A DC	. €4.60	
5A DC	€4.60	
20V DC	€4.60	100
50V DC	. €4.60	100
300V DC	€4.60	- 800
300 V AC	€4.75	1.50
VU Meter	€5.95	_



#### CLEAR PLASTIC MODEL MR 52P

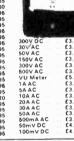
LEANI			16
Size: 60 x	60n	m	
50u A			€4.10
100uA			€3.8
500 u A			£3.7
50-0-50u			£3.8
100-0-10	0uA		€3.7
1 m A			€3.6
5mA			€3.6
10mA			€3.6
50 m A			€3.5
100mA			€3.6
500m A			€3.5
1 A DC			£3.6
5A DC			€3.6
10V DC			£3.6
20V DC			€3.6
50V DC			€3.6
300 V D C			€3.6
15V AC			
300 V A C			€3.7



Calling	-
5 Meter 1n	nA. £4.20
VU Meter	£4.85
1A AC	£3.65
5A AC	€3.65
10AAC	€3.65
20A A C .	€3.65
30 A A C	£3.65

#### DEL MR 65 Size: 80 x 80mm

BAKELI	TΕ	M	O C	Eι
25uA			€5.	80
50uA			£4.	
100uA			£4.	
500uA			£4.	
50-0-50u			£4.	
100-0-100			£4.	
600-0-500	)uA		€3.	
1mA			£3.	
1-0-1 mA			€3.	
5mA			€3.	
10mA			£3.	
50 m A			€3.	
100mA			£3.	
500mA			£3.	
1 A DC			€3.	
2A DC			£3.	
5A DC			€3.	
10A DC			£3.	95
15A DC			€3.	
30A DC			€3	
50A DC			£4.	
5V DC			£3.	
10V DC			€3.	
15 V D C			€3.	
20 V D C			£3.	
50 V D C			£3.	
150V DC			€3.	95



#### CLEAR PLASTIC MODEL MR 65P

50uA. (4.35) 200uA (4.25) 200uA (4.25) 200uA (4.25) 200uA (4.25) 100.0-100uA. (4.20) 100.0-100uA. (4.20) 100.0-100uA. (4.20) 100.0-100uA. (4.20) 100.0-100uA. (4.10) 100uA. (4.10) 100uA.	iza: Bo x / bmm	l .		
200 A		£4.35	(	
500uA	100 u A			
50.0 - 500 A	200u A	€4.20	ALTERNATION OF THE PARTY OF THE	. 1
100-0-100UA . £4.20 500-0-500UA £4.10 1mA £4.10 50mA £4.10 10mA £4.10 50mA £4.10 50mB £4.10	500uA	£4.15	A	0,
500-0-500-A	50-0-50uA	£4.25	RA	- 1
1mA	100-0-100uA	£4.20		
1-0-1mA	500-0-500uA	£4.10		
5mm         £4.10         300V DC         £4.10           10mA         £4.10         15V DC         £4.20           50mA         £4.10         50V AC         £4.20           50mA         £4.10         50V AC         £4.20           500mA         £4.10         300V AC         £4.20           500mA         £4.10         300V AC         £4.20           5ADC         £4.10         5 Meter ImA         £4.82           15ADC         £4.10         VMeter         £8.20           15ADC         £4.10         AV Meter         £4.20           15ADC         £4.10         AV Meter         £4.20           30ADC         £4.10         AV Meter         £4.10           50ADC         £4.10         20AAC         £4.10           50ADC         £4.10         20AAC         £4.10           50ADC         £4.10         20AAC         £4.10           15V DC         £4.10         50MAC         £4.10           15V DC         £4.10         50MAC         £4.10           20V DC         £4.10         50MAC         £4.10           50V DC         £6.10         20MAC         £4.10	1mA	£4.10		
10mA	1-0-1mA	€4.10	1 - 14 - 161	- Lil.
SOMA	5mA	£4.10		
100mA	10mA .			
\$500mA	50mA .			
TADC         C410         500V AC         R4 30           SADC         C410         500V Befor ImA         C4 30           SADC         C410         VW Meter         C5 20           15A DC         C410         VU Meter         C5 20           15A DC         C410         1A AC         C4.10           20A DC         C420         5A AC         C4.10           30A DC         C4.25         10A AC         C4.10           50A DC         C4.10         20A AC         C4.10           10V DC         C4.10         50A AC         C4.10           15V DC         C4.10         50M AC         C4.10           20V DC         C4.10         100m AC         C4.10           50V DC         C4.10         20m AC         C4.10	100mA:	£4.10		
\$A.DC	500mA			
10A DC				
15A DC	5A DC	€4.10		
20A DC	10A DC .	£4.10	VU Meter	
30A DC         £4.25         10A AC         £4.10           50A DC         £4.10         20A AC         £4.10           5V DC         £4.10         30A AC         £4.10           10V DC         £4.10         50A AC         £4.10           15V DC         £4.10         50A AC         £4.10           20V DC         £4.10         100mA AC         £4.10           50V DC         £4.10         200mA AC         £4.10	15A DC	£4.10		
50A D C £4.10 20A A C £4.10 5 V D C £4.10 30A A C £4.10 10 V D C £4.10 50M A C £4.10 15 V D C £4.10 50M A C £4.10 20 V D C £4.10 100M A C £4.10 50 V D C £4.10 200M A C £4.10	20A DC	€4.20	5A A C	
5VDC	30 A D C			
10V DC £4.10 50A AC £4.10 15V DC £4.10 50M AC £4.10 20V DC £4.10 100M AC £4.10 50V DC £4.10 200M A AC £4.10				
15V DC £4.10 50mA AC £4.10 20V DC £4.10 100mA AC £4.10 50V DC £4.10 200mA AC £4.10	5V DC	£4.10	30A A C	
20V DC £4.10 100mA AC £4.10 50V DC £4.10 200mA AC £4.10	10V DC	€4.10	50A AC	
50V DC . £4.10 200mA AC . £4.10	15V DC	£4.10	50mAAC.	
	20V DC			
150V DC £4.10 600mA AC £4.10	50V DC			
	150V DC	£4.10	500mA AC	£4.10

240° Wide Angle 1mA METERS MW1-6 60 x 60 mm

£7.00 P/P & Ins 15p MW1-8,80 x 80 mm £7.45P& Pins.15p



#### RATTERY/LEVEL PANEL INDICATOR

18mm Panel mounting OUR PRICE 95p P/F



R Ins 15p Discounts for quantity



VU METER TYPE 3 Size: 33mm > 20mm. £1.55 P/P & Ins 15p

#### REDUCED TO CLEAR

YAMABISHI VARIABLE **VOLTAGE TRANSFORMERS** 

VOLTAGE TRANSFORMERS Excellent quality at low cost. Input: 230V 50/60Hz, Output 0—260V. MODEL S260 BENCH MOUNTING P/P & Ins 1A E7, 30-E1.50 2.5A £10.35-£1.50 BA £19.45-£1.50 10A £22.75-£1.50 10A £22.75-£1.50 10A £23.75-£1.50 55A £41.95-£1.50 40A £73.40-£1.50



#### make sense of Test Equipment

#### AUDIOTRONIC Model ATM1

AUDITOT HUNK! Top value 1,000 opv pocket multi-meter. Ranges: - 0/10/50/250/1,000 volt AC and DC. DC current 0-1mA/100mA. Resistence: 0/150k ohms. Decibels: - 10 to +22dB, Size 90 x 60 x 28mm. Complete with test leads.



OUR PRICE £4.25 P/P & Ins 25p

#### AUDIOTRONIC MODEL ATMS

Javel movement, attractively moulded case with edgwise ohms adjustment. Ranges: 0-3/15/150/300/1200 AC, (2500 opv), 0-6/30/300/500 DC, (2500 opv), 0-300 uA/0-300mA DC. Resistance: v 10 & Resistance: v 10 Resistance: x 10 & x 100. — 10 to +16dB. Supplied with battery test leads and data booklet. Size: 121 x 73 x 29mm.

OUR PRICE £4.85 P/P & Ins 25p

#### MODEL C7202EN

20,000 o.p.v. DC.
10,000 o.p.v. AC
Mirror Scale.
5/25/60/250/500/
1000/2500 V. DC.
10/50/100/500/1000
V. AC. DC Resistance.
x10, x1000 (300)
centre scale) DC
Current 50uA/
25mA/250mA/.—20
to +88 dB.



OUR PRICE £7.50 P/P & Ins 30p

BARCLAYCARD & ACCESS Phone your order to 01-200 1321 or call into any branch





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# IT MAKES SENSE TO

#### HIOKI 730X

30,000 opv. Over-load protection. 60/µA/ 30mA/300mA. 2K/200K

2 Meg Ohm. -10 to +63 dB OUR PRICE £8.10



#### **U4323 MULTIMETER**

U4323 MULTIME 20,000 pv. Simple unit with audio/IF oscillator. Suitable for general receiver tuning. Ranges: 0.5/2,5/10/50/250/ 500/1000V DC. 2.5/40/5/5/06/500/



2.5/10/15/250/500/1000V AC. 0.05/ 0.5/5/50/500mA DC. Resistance: 0.5/6/50/500mA DC. Resistence: x10, x100, x1,000, x10,000 (50Ω. 500Ω δkΩ.50kΩ bentre scale) Bettery operated. Size: 160 x 97 x 40mm. Supplied in carrying case com-lets with test leads. OUR PRICE FR 60 P/P & Ins 60p

#### TMK 200 MULTIMETER KIT

multimeter and save money. Complete kit with mater scale. movement and

movement and rotary range selector ready mounted in cabinet. All parts, batteries, test prods and instructions. Ranges: 0/0.6/6/30/120/600/1200V D.C. 0/6/30/120 600/1200 V A. C. Current: 0/0 6/6/ 60/600mA. Resistance: 0/10 100K/1/10 Meg ohms. Decibels —20 to #63db. Size: 90 x 150 x

OUR PRICE £9.65 P/P & Ins 30p

#### MODEL C7208FM

30,000 opv DC 15,000 opv AC 6/3/15/60/300/600/ 1200 V. D.C. 6/30/ 120/600/1200 V. A.C. 120/600/1200 V. AC DC Resistance x1, x10, x100, x1000 (50Ω centre scale) DC Current 30uA/ 3/30/600mA. —20 to +63dB



OUR PRICE £9.65 P/P & Ins 30p

#### 114324 MILL TIMETER fligh sensitivity,

OUR PRICE £10.60P/P & Ins 60p

#### U91 Clamp VOLT AMMETER

ANIMETER
For measuring AC voltage and current without breaking circuit. Ranges: 300/600V AC, Current: 10/25/100/250/500A. Accuracy 4%, Size 283 x 94 x 36:nm, Complete with carrying case, leads and fuses.

OUR PRICE £15.10 P/P & Ins 60p OUR PRICE £18.35 P/P & Ins 60p

#### **U4312 MULTIMETER**

U4312 MULTIM extremely sturely instrument for general electrical use. 667 pc; 0/0.3/1.5/7.5/30/ 60/150/300/600/ 900V DC & 75mV 0/0.3/1.5/7.5/30/ 60/150/300/600/ 900V AC. 0/300µ 0/1.5/6/15/150/60/ 00mA/11.5/6A DC. 0/1.5/6/15/

DC. 9/1.5/6/15/ 60/150/600mA/ 1.5/6A AC. 0/200/3k/30k ohms. DC accuracy 1%. AC 1.5%. Knife edge pointer, mirror scale. Complete with sturdy metal carrying case, leads and instructions.

OUR PRICE £11.60P/P & Ins 60p

#### HIOKI 750X VOLT-OHM-

HIO KI 750X VOLT-OHMMILLIAMETER
43 ranges: 0-0.3/0.6/
1.5/3/6/12/30/60/150/
300/600/1.200V AC.
Current: 0-30/600.4/
1.5/3/15/30/60/120/
300/600/1.200V AC.
Current: 0-30/600.4/
1.5/3/15/30/60/120/300/
Decibels: -10 to +17dB. Output:0-3/6/15/30/60/120/300V. Accursev ± 3% DC. ± 4% AC. Sensitivity:
50,000 opv DC. 5,000 opv AC. 4 inch
meter. Built in protection. Size: 57 x
102 x 153mm.
DUR PRICE £12.90P/P.R. Ins. 600

OUR PRICE £12.90P/P & Ins 60p

#### TMK MODEL TW50K

TMK MODEL TWBUK
46 ranges, mirror
scale, 50k/V DC
50k/V AC,
DC Volts: 0.125/
0.25/1.25/2.5/5/10/
25/50/125/250/
100/2000, AC Volts
1.5/3/5/10/25/50/
100/2005/00mA/5/
10A, Resistence:
10k/100k/1 Meg/
10 Meg ohms. -20 to +81.5dB.

OUR PRICE £13.50P/P & Ins 60p

#### MODEL AF. 105 VOM

protection. 0/-3/3/12/60/120/ 300/600/1200V D.C. 0/6/30/120/ 300/600/1200V DC 0/30μA/6/ 60/300 mA/ 12 Amp. 0/10K/ 1m/10m/100

OUR PRICE £13.50P/P & Ins 60c

#### MODEL 500 30,000 opv with overload protect-tion. Mirror scale. 0/0.5/2.5/10/25/

0/0.5/2.5/10/25/ 100/250/500/ 1000V DC. 0/2.5/10/25/100/ 250/500/1000V AC. 0/50u A/5/50/ 500mA. 12A DC. 0/60k/6 meg/60 m

OUR PRICE £15.05P/P & Ins 60p

#### **U4317 MULTIMETER**

High sensitivity instrument for field and laboratory work. Knife edge pointer, 86mm. mirror scale. Overload protection.

Overload protection.
Ranges: 100mV/
0.5/2.5/10/25/50/100/250/500/1000
V.D. 0.5/2.5/10/25/50/100/25/50/100/25/50/100/25/50/100/25/50/100/25/50/100/25/50/100/25/50/25/60/

#### MODEL AS 1000 VOM

100,000 opv. Mirror scale Built-in meter protection. 0/3/ 12/60/120/300/ 600/1200V DC 600/1200V DC. 0/6/30/120/300/ 600V AC. 0/10µA/ 6/60/300mA/ 12 Amp. 0/2K/ 200K/2M/200 Meg Ohm. – 2010 117

DUR PRICE£18.90P/P & Ins 60p

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#### KAMODEN 360 MULTIMETER

High sensitivity. DC 100kohm/V AC 10kohm/V 5" mirror scale. AC 10kohm/V
AC 10kohm/V
5" mirror scale
overload protect
ed. Ranges: 0.5
2.5/10/50/250/
1000V DC. 5/10/
50/250/1000V
AC. Current:
0.01mA/0.5/5/50/
500mA/10A.
Resistance: 0.1/
1/10/100 ohms/
10/100M ohms/
10/100M ohms/
10/100M ohms/ -20 to

Decibels —20 to +62dB, Battery operated, Size: 180 x 140 x 80mm, Supplied complete with,

00

OUR PRICE £18.90 P/P & Ins 60p

#### Model HT100B4 MULTIMETER

Model HT100BA MULTIMETER
Overload protected,
shock proof circuits,
9,5uA Meter with
mirror scale. Sensitivity,
100k. V. Polarity change
switch. Ranges: 0.5/2.5/1
1/50/250/500/1,000
Volts DC. 2.5/10/50/
200k/2/20 Meg. ohms.
DC current: — 0.10A.—20
to +62dB.—0ptrate: from 2 x 1,5V
batteries. Size: 180 x 134 x 79mm.

OUR PRICE £21.50P/P & Ins 60p

#### 370WTR MULTIMETER

37/WH M MULTIME I Features AC current ranges. 20,000 pp. 0/0.5/2.5/10/50/ 250/500/1000 V DC. 0/2.5/10/50/250/ 500/1000 V AC. 0/50uA/11/01/00 mA/11/04 DC. 0/100mA/11/04 AC. 0/5k/50k/500k/ 5 Meg/50 Meg. Decibels: -20 to +6 2dB.

OUR PRICE £21.50P/P & Ins 60p

#### MODEL C7080EN

Giant 6" mirror pcale, 20,000 opv, 0/0.25/1/2.5/10/ 50/250/1000/ 5000V DC, 0/2.5/10/50/250/ 1000/5000V AC. 0/50uA/1/10/ 100/500mA/10 DC. 0/2k/200k/ 20 Meg. —20 to A/10A

OUR PRICE£21.50P/P&Ins 60p

#### KAMODEN 72.200 Multitester

NAMUJEN 72. High sensitivity tester, 220,000 opp Overload protected. Micror scale. Micror scale. Micror scale. Micror scale. Micror scale. 200,061,33(0)120/500/120/500/120/500/120/500/120/500/41200 VAC. 0/680A/1,2mA/120mA/500mA/12A DC 0/12A AC. —20 to 453dB. 0/26/200k/2 Meg/200 Megodom

OUR PRICE £24.30P/P & Ins 60p

#### KAMODEN HM7208 FET VOM

B

. .

P/P &

NAMUUEN H1872
Input impedence 10
Megohms. Ranges:—
0/.25/1/2.5/10/50/
1000V DC. 0/2.5/10
50/250/1000V AC.
0/25u/4/2.5/25/250
mA DC.
0/5k/50k/500k/5 M
500 Megohms OUR PRICE

£24.30 P/P & Ins 60p

#### SWR METER Model SWR3

Handy SWR meter for transmitter antenna alignment, with built-in field strength meter. Accuracy 5%, Impedence 52 Indicator 100uA DC, Full scale 5 section collapsible antenna. Size 145 x 50 x 60mm.

OUR PRICE £4.55 P/P & Ins 60p

#### U4341 Multimeter &

Transistor Tester 7 ranes. 16,700opv. 8 ranes. 16,700opv. 8 ranes. 16,700opv. 16,700 Transistor Tester Resistance: 0.06/ 0.6/2/6/20/60/200k ohms/2 Mohms

Battery operated. Supplied complete with probes, leads and steel carrying case. Size: 115 x 215 x 90mm.

OUR PRICE £11.85 P/P & Ins 60p

#### KAMOBEN TT35

KAMODEN 1T35
TRANSISTOR TESTER
High quality
instrument to
test reverse leak
current and DC
current. Amplification factor of
NPN, PNP, Giodes,
transistors, SCR's
clear scale merclear scale m

OUR PRICE £18 90

#### S100TR MULTIMETER

TRANSISTOR TESTER TRANSISTOR TESTER
100,000pp. Mirror
sale, Overload
protection, 0/0,12/
0.6/3/12/30/120/
120/600V AC.
0/12/600V AC.
0/12/600V AC.
0/10/6/18/00/
120/600V AC.
0/10/6/18/00/
120/600V AC.
0/10/6/18/00/
120/600V AC.
0/12/600V AC.
0/

OUR PRICE £22.65P/P & Ins 60p

#### LB4 TRANSISTOR TESTER

Tests PNP or NPN transistors. Audio indication. Operates on two 1.5V batteries. Complete with instructions etc OUR PRICE £4.85 P/P & Ins 20p

O.

CIS PULSE OSCILLOSCOPE

C15 PULSE OSCILLOSCOPE
For display of pulsed
and periodic wave
forms in electronic
circuis, VERT. AMP.
Bandwidth: 10MHz.
Sensitivity at 100kHz
VRMS/mm: 0.1—25
HOR. AMP. Band
width: 500kHz
Sansitivity ay 100kHz
Frest triggered recep
1—3000usec. Free trunning 20—200
kHz in nine ranges. Calibrator pips.
220 x 360 x 430mm. 115—230V AC.
HIR PRICEF647 56 P/P & Ins.

OUR PRICE £47.50

#### RUSSIAN CI16 Double Beam

OSCILLOSCOPE OSCIELUSCUTE
5 MHz pass band.
5 mHz pass band.
5 separate Y1 and Y2
amplifiers. Rectangular 5" x 4" CRT.
Calibrated triggered sweep from 0.2usec.
to 100 milli-sec/cm.
Free running time Free running time base, 50Hz-1MHz. Built-in time base

Calibrator and amplitude Calibrator, supplied complete with all accessories and instruction manual OUR PRICE £93.95

MDDEL TE15 GRID DIP METER

GRID DIP METER Transiatorised. Oper-ates as Grid Dip, Oscillator, Absorb-tion Wave Meter and Oscillating Detector. Frequency range 440kHz – 280MHz in six coils. 500uA meter, 9V bettery operation. Size: 180 x 80 x 40mm.



OUR PRICE £18.90 Ins 30p

Tech O

#### SINCLAIR DM2 DIGITAL MULTIMETER



Will measure AC and DC volts, AC and DC current, and resistance in a total of 20 ranges. The large light emitting diode display will read up to 1999 and automatically indicate polarity. Indication of positive and polarity. Indication of positive and negative overload is also provided The instrument is fitted with a combined carrying handle and bench stand and sockets are provided for the connection of an external power supply. RANGES:

DC VOLTS: 1v, 10, 100v, 1000v, AC VOLTS: 1v, 10v, 100v, 1000v, DC CURRENT: 1mA, 10mA,

100mA, 1000mA

AC CURRENT: (1mA.10mA), 100mA, 1000mA RESISTANCE: 1k, 10k, 100k, 1000k OUR PRICE £63.70P/P & Ins 50p

#### TRANSISTORISED L.C.R. A.C.



RANISTORISED L.C.R. A.C.
BR/8 MEASURING BRIDGE
Anaw portable bridge offering excelent range and accuracy at low cost. Resistence: 6 ranges: 0.1 chm-11.1 megohm ± 1% Inductance: 6 ranges: 1 microhenry-111\* herries ± 2% Capacity: 6 ranges: 10gf-1110 mfd ± 2% Turns Ratio: 6 ranges: 11/1000-1111100 ± 1% Bridge Voltage at 1,000cps. Operated from 9-volt battery. 100 micro-amp meter indication. Size 7½ x amp meter indication. Size 71" 5" × 2" OUR PRICE £29.70 Ins our

#### TE-200 RE SIGNAL GENERATOR

Accurate wide range Accurate wide range signal generator covering 120 kHz-500 MHz on 6 bands. Directly calibrated. Variable R.F.

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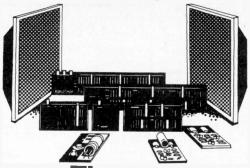
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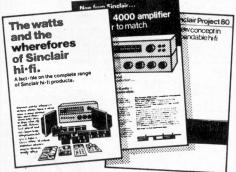
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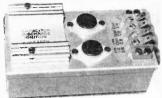
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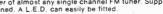
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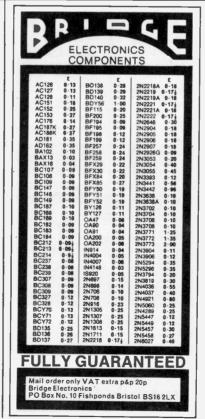
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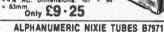
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## INDEX TO ADVERTISERS

Idea Hell (D.E. Conn. Boo)	#1.1	
dam Hall (P.E. Supplies)	Elvins762	Osmabet700
stro Electronics	Euray Trading754	
utomated Homes771		Phoenix Electronics Ltd762
xial Products768	Fibre Optics770	Phonosonics759
	Flairline Supplies726	Phototech (Europe)769
Balmaha	Foulsham-Tab Ltd758	Phototech (Europe) 769 Pulse Electronics Ltd. 69
Barrie Electronics758	G.P. Electronics769	Radio Components Specialists702-703
Beta Devices768	Greenweld Electronics760, 767	Radio Exchange 699
3i-Pak708-709	G.W.M. Radio	Radio Shop, The
3i-Pre-Pak Ltd	G. W. W. Madio	Radnage77
lib Hi-Fi Ltd694	Hamman Complete Co. 144	Rean Products
loffin Projects769	Harverson Surplus Co. Ltd	Reedhampton Ltd
ridge Electronics766	Heath (Gloucester) Ltd710	P & TV Components Ltd 606 60
Iritish Institute of Engineering	Henry's Radio761	R & TV Components Ltd
Technology730, 764		R.T. Services
ritish National Radio & Electronics	I.L.P. Electronics Ltd707	n.i. 38141083/00
School706, 725	International Electronics Unlimited729	Colon Catavariana 37
uckingham Electronics	Island Devices769	Salop Enterprises
ywood704		
ywood	Jones, J. C768	Scientific Wine Co77
ambridge Learning Enterprises 767	J.W.B. Radio	Scott Electronics704
ambridge Learning Enterprises767 hiltmead Ltd730		Selektron
hromasonic Electronics	Kensington Supplies769	Sinclair (Hi-Fi)704
lef Products	Kensington Supplies/69	Sinclair (IC20)698
	L no locio 200 200 200 200 200 200 200 200 200 20	Sinclair (Project 80)760
opper Supplies	Lasky s755-756-757	Sinclair (System 4000)752
rescent Radio Ltd764	London Computer Operators Training Centre769	Sintel760
T. Electronicscover iii	Centre	Stevens, P. F77
avian698		Technalogics77
.E.W. Ltd	Maplin Electronic Suppliescover iv and 770	Technomatic Ltd
ziubas, M771	Marco Trading768	Teleradio Electronics
	Marshall, A., & Son765	Trafalgar Supplies700
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AD162	45p	BD123	85p	OA47	10p	ZTX500	18p	2N3709	14p		— 1	MFC4000	250mW
AF114	25p	BD124 BD131	70p	OA70	10p	ZTX501	20p	2N3771	£1-50	BRIDGE		Audio	7
AF115 AF116	25p 25p	BD132	40p 50p	OA79 OA81	10p 10p	ZTX504 ZTX531	50p 30p	2N3772 2N3792	£1-60 £1-00	RECTIFIERS W02 1A 200V	38p	TBA800 5 Audio	watt 9
AF117	25p	BD135	50p	OA90	10p	ZTX550	25p	2N3794	30p	BY164 1 4A	· 1		Amp D.I.L
AF118 AF124	50p 30p	BD136 BD139	50p 75p	OA91 OA200	10p 10p	IN659	8p	2N3819 2N3820	35p 55p	200V MDA952 2 6A	57p	TO99	4
AF239	50p	BD140	87p	OA202	10p	IN914 IN916	8p 8p	2N3823	70p	100V	80p	741C Op .	Amp 8 14 1099 3
BA102	30p	BD153 BD156	80p 80p	OA210	35p	IN4001	8p	2N3866	35p		_	748C Op :	
BA112 BA114	50p 15p	BDY17	£1·50	OA211 OC16	35p 90p	IN4002	9p	2N3904 2N3905	22p 25p	ZENER DIODE		DIL	7
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BA156	15p	BDY19 BDY20	£1-90 £1-20	OC22	55p	IN4004	10p	2N4014 2N4036	30p 62p	1 5W range	25p	Amp	11ء - dio I.C. £1
BC107 BC108	13p 12p	BF152	20p	OC26 OC28	65p 60p	IN4005 IN4006	12p	2N4036 2N4037	44p	10W range	45p	TAD100 R	
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BC172	13p	BFY52	25p	TIP30	55p	2N1711 2N1890	30p 60p	2N5195	£1-40	21 × 31	,	32p	23
BC182 BC182L	12p	BFY53 BFY90	25p 65p	TIP31A TIP32A	57p 69p	2N2145	15p	2N5245 2N5296	45p 55p	21 × 31 21 × 5		35p	350
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BC185	35p	BSX20 BSX21	15p 25p	TIP41A	7 <b>0</b> p	2N2222	20p	2N5485	55p	PIN INS. TOOL		72p	72p
BC186 BC212	27p 12p	BY127	20p	TIP42A TIP29B	85p 54p	2N2222A 2N2306	25p	2N5490 2N5555	55p 65p	SP.F. CUTTER 100 PINS SS		52p 30p	52g 30g
BC212L	140	BY164	65p	TIP30B	60p	2N2306 2N2369A	70p 50p	2N5777	40p	100 PINS DS		30p	30g
BC213	12p	IS100	15p	TIP31B	65o	2N2477	30p	PERSO	NAL	500 PINS SS		- 55	£1 5
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BC214L	15p	MJ481	95p	TIP34B	£1-62	2N4904	25p	ALW					
		QTY DI	SCOUNT	S 12 + 109	6 25 +	15% 100 +	20%	WELC	OME	ALSO ST	COC	KED	
									$\neg$	BNC PLUGS at 3		KLD	
TTL (ful				SN7433	70p	SN4753	20p	SN7451,	20p	UHF (N) PLUGS &	it 50p ea	ach.	
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SN7403	20p	SN7417	30p	SN7441AN	75p	SN7472	30p	SN7490	75p	All the above an Please add 8% V	e new ATP&I	m origina 230p.	ıı packets
SN7404 SN7405	20p 20p	SN7420 SN7422	20p 38p	SN7442 SN7443	75p £1:00	SN7473 SN7474	40p 48p	SN7491A	N ε1·00				
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SN7407	30p	SN7425 SN7427	38p	SN7446	€2.00	SN7476 SN7480	45p 80p	SN7493 SN7494	75p	Potention			
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SN7410	13p	SN7430	20p	SN7450	20p	SN7482	87p	SN7496	£1-00	Rotary Pots Rotary Switched	17 25	p D	45p
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2} × 3}		0 · 1 32p	23p
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2½ × 5 3½ × 3½		0·1 32p 35p 35p	23p 35p 35p
2½ × 5 3½ × 3½ 3½ × 5		0·1 32p 35p 35p 40p	23p 35p 35p 41p
2½ × 5 3½ × 3½ 3½ × 5 17 × 2½		0·1 32p 35p 35p 40p £1·05	23p 35p 35p 41p 90p
2½ × 5 3½ × 3½ 3½ × 5 17 × 2½ 17 × 3½		0·1 32p 35p 35p 40p £1·05 £1·43	23p 35p 35p 41p
2½ × 5 3½ × 3½ 3½ × 5 17 × 2½		0·1 32p 35p 35p 40p £1·05 £1·43 £1·84	23p 35p 35p 41p 90p £1 · 12
2½ × 5 3½ × 3½ 3½ × 5 17 × 2½ 17 × 3½ 17 × 5		0·1 32p 35p 35p 40p £1·05 £1·43	23p 35p 35p 41p 90p
2½ × 5 3½ × 3½ 3½ × 5 17 × 2½ 17 × 3½ 17 × 5 PIN INS. TOOL SP.F. CUTTER 100 PINS SS		0·1 32p 35p 35p 40p £1·43 £1·84 72p 52p 30p	23p 35p 35p 41p 90p £1·12 72p 52p 30p
2½ × 5 3½ × 5 17 × 2½ 17 × 3½ 17 × 5 PIN INS. TOOL SP.F. CUTTER 100 PINS SS 100 PINS DS		0 · 1 32p 35p 35p 40p 40p 51 · 05 £1 · 43 £1 · 84 72p 52p 30p	23p 35p 35p 41p 90p £1·12 72p 52p 30p 30p
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2½ × 5 3½ × 3½ 3½ × 5 17 × 2½ 17 × 3½ 17 × 5 PIN INS. TOOL SP.F. CUTTER 100 PINS SS 100 PINS SS 500 PINS SS	!	0·1 32p 35p 35p 40p 51·05 £1·43 £1·84 72p 30p £1·55	23p 35p 35p 35p 41p 90p £1·12 72p 52p 30p 30p £1·55

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