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27-23 P.P. & INS. 44p (Overseas P.P.

NEW EDU: KIT MAJOR

COMPLETELY SOLDERLESS ELECTRONIC CONSTRUCTION KIT BUILD THESE PROJECTS WITHOUT SOLDERING IRON OR SOLDER

- 4 Transistor Earpiece Radio
- Signal Tracer
 Signal Injector
 Transistor Te Tester NPN
- PNP Transistor Push Pull
- Amplifier

 5 Transistor Push Pull
 Amplifier
- o 7 Transistor Loud-speaker Radio MW/LW. o 5 Transistor Short Wave Radio
- Short
- Electronic Metronome
 Electronic Noise Genera-Batteryless Crystal Radio.
 One Transistor Radio.
- Transistor Regenerative Radio
 Transistor Regenerative Radio
 Continuity
- Continuity
- Audible Continui
 Tester
 Sensitive Pre-Amplifier

Components include:

24 Resistors ● 21 Capacitors ● 10 Transistors ● 3½ Loudspeaker
 3 12-way Connectors ● 2 Volume Controls ● 2 Slider Switches ● 1 Tuning Condenser ● 3 Knobs
 Ready Wound MW/LW/SW Cols ● Ferrite Rol ● 6½ yards of wire ● 1 yard of sleeving, etc.
 Parts price list and plans 50p (free with parts)

NEW ROAMER 4 NINE

WITH V.H.F. INCLUDING AIRCRAFT

sistors. Tunable wave

bands as Roamer Ten, Built in ferrite rod aerial for MW/LW

in ferrite rod aerial for MWLW. Retractable chrome plated telescopic aerial for VHF and SW. Push Push Pull output using 600 nW transistors. 9 Transistors and 3 diodes, tuning condenser with VHF section, separate coil for aircraft, moving coil outlapsaker, volume ON/OFF and wavechange controls. Attractive all white case with red grille and carrying strap. Size 94m × 7m × 22m approx. Parts price list and plans 30p (FREE with parts).

TOTAL BUILDING £6-95 COSTS (+10% VAT 69p)

P.P. & INS. 44p (OVERSEAS P. & P. £1 85)

Total Building Costs

(+10% VAT 22p) P.P. & ins. 26p

POCKET FIVE

Tunable wavebands MW/LW and Band 7 st Trawler Sand. 7 stages, ransistors and

transistors and 2
diodes, supersensitive ferrite
rod aerial, moving coit
loudspeaker, attractive
Black and (fold Case. Size
5\(\frac{1}{2}\)in x 1\(\frac{1}{2}\)in x 2\(\frac{1}{2}\)in
Plans and parts price list 15p
(FREE with parts). (Overseas P. P. £1 25)



Total Building Costs £2.50 P.P. & Ins.

(+10% VAT 25p) (Overseas P. & P. £1:25)

TRANSONA

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NEW EVERYDAY SERIES



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EV5 5 Transistors and 2 diodes. MW/LW. Powered by 44 volt Battery. Ferrite rod aerial, tuning condenser, volume control, and loud-speaker. Attractive case with red speaker grille. Size 9in × 54in × 24in approx. Parts price list and plans 15p (FREE with parts).

TOTAL BUILDING COSTS

P.P. & INS. 30p (OVERSEAS COSTS)

P.P. & INS. 30p (OVERSEAS COSTS)

EV6 Case and looks as above.
6 Transistors and 3 diodes. Powered by 9 volt Battery. Ferrite rod aerial, 3' loudes, Powered by y
volt Battery. Ferrite rod aerial, 3' loudespeaker, etc.,
MW/LW coverage, Push Pull Output. Parts price
list and plans 15p (FREE with parts.)
TOTAL
BUILDING
COSTS

P.P. & INS. 30p
(OVERSEAS
P. & P. £1-25)

EV7 Case and looks as above Costs

Case and looks as above.
Transistors and 3 diodes.

Six wavebands.

MW/LW, Transistors and 3 diodes.

Six wavebands.

S

TOTAL BUILDING £4-08 (+10% VAT 40p)

ROAMER EIGHT Mk. I

HTIW WOR **VARIABLE** TONE CONTROL

7 TUNABLE WAVEBANDS

7 TUNABLE WAVEBANDS:
MW1, MW2, LW, SW1
SW2, SW3 AND TRAWLER BAND. Built-in ferrite rod
aerial for MW and LW. Retractable chrome plated telescopic aerial for short waves. Posh-pull output using
600m W transistors. Car aerial and tape record sockets.
Selectivity switch. 8 transistors plus 3 diodes. Latest 4"
watt Ferrite Magnet bundspeaker, Air spaced ganged
tuning condenser. Volume/on/off, tuning, wave change
and tone controls. Attractive cases in rich chestnut shade
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and plans 25p (FREE with parts).

TOTAL BUILDING £6-98 (+10% VAT 69p)

(OVERSEAS P. & P. £1-85)

ROAMER WITH VHF INCLUDING

10 TRANSISTORS 9 TUNABLE WAVEBANDS.

AIRCRAFT



TOTAL BUILDING

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P.P. & INS. 52p (OVERSEAS P. & P. £1-85)



Components include: Tuning Condenser: 2 Volume Controls: 2 Slider Switches: Fine tone 3" moving coil Speaker: Terminal Strip: Ferrite Rod Aerial: Battery Clips: 4 Tag Boards: 10 Transistors: 4 Diodes: Resistors: Capaciters: Three Jin Knobs, Units once constructed are detachable from Master Unit, enabling them to be stored for future use. Ideal for Schools, Educational Authorities and all those interested in radio construction. Parts price list and plans 25p

TOTAL BUILDING

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P.P. & INS. 33p (OVERSEAS P. & P. £1-85) . & P. £1 85)

TRANS EIGHT 8 TRANSISTORS AND 3 DIODES

TURABLE WAYEBANDS, MW, LW, SWI, SW2, SW3 AND TRAWLER BAND. Sensitive ferrite rod aerial for MW, and LW. Telescopic aerial for short waves. 3in speaker. 8 improved type transistors plus 3 diodes. Attractive case in black with red grille, dial and black knobs with polished metal inserts.

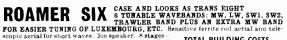
Size 9in × 54in × 24in approx. Pushpull output. Batter economiser

fin a back output. Back for extended frive a Battery economiser witch battery Ample power to drive a larger speaker Parts. price list and plans 25p (FREE with parts).

TOTAL BUILDING COSTS

£4-48 P.P. & INS. 33p (OVERSEAS P. & P. £1-25)

(+10% VAT 44p)



POR EASIER TUNING OP LUXEMBOURG, ETC. Sensitive letter to the scopic aerial for short waves. 3in speaker. Stateges 6 transistors and 2 diodes, etc. Attractive black case with red grille, dial and black knobs with polished metal inserts. Size 9in ×51in ×21in approx.

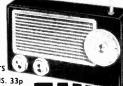
Plans and parts price list 25p (FREE with parts).

TOTAL BUILDING COSTS

TOTAL BUILDING COSTS

(OVERSEAS P. & P. & P. & IN. 31p (OVERSEAS P. & P. & P. & I. 85)

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VOLUME 10 No. 7 JULY 1974

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NIM MACHINE by R. D. Mount An unbeatable opponent for the old chinese stick game	627
GENERAL FEATURES FIRST STEPS IN CIRCUIT DESIGN—4 by A. P. Stephenson Biasing gain and the emitter follower	596
INGENUITY UNLIMITED Synthesiser patchboard	603
SOLAR POWER AND ITS APPLICATIONS FOR TODAY by P. S. Woodcock A look at the utilisation of solar energy and its effects	61 8
NEWS AND COMMENT EDITORIAL—SOUND POWER	589
SPACEWATCH by Frank W. Hyde The Sun by Skylab	595
ELECTROMUSE by Malcolm Pointon A musician discusses the application of electronic techniques to composition and performance	600
BOOK REVIEWS Selected new books we have received 60	04, 616
NEWS BRIEFS Marconi Centenary Exhibition—Obituary 61	6, 625
TUNING IN TO ESP by Peter Verwig A report from a recent IEEE conference in America on parapsychology	617
INDUSTRY NOTEBOOK by Nexus What's happening inside industry	622
PATENTS REVIEW Reminders on general British Patent procedures	631
READOUT A selection of readers' letters	634
SOUND SYNTHESIS FOR THE AMATEUR A special offer to Clubs, Schools and Universities to hear the P.E. Lecture on "Sound Synthesis" using the P.E. Synthesiser	635
Our August issue will be published on Friday, July 12, 1974	

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1/2	2%	$\Omega MI = \Omega \Omega$	E24	3-5p	3p
ł	10%	1 Ω = 3 · 9 Ω	EI2	1-3p	[-lp]
å	5%	4 7 Ω – I M Ω	EI2	1-3p	l-lp
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.0.5 watt 5% Iskra resistors 5 off each value 4.7 Ω to IM Ω E12 pack 325 resistors £2.40. E24 pack 650 resistors £4.70

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Carbon track $5k\Omega$ to $2M\Omega$, log or linear (log $\frac{1}{4}W$, lin $\frac{1}{2}W$). Single, 14p. Dual gang (stereo), 49p. Single D.P. switch 28p.

SKELETON PRESET POTENTIOMETERS. Linear: $100, 250, 500 \Omega$ and decades to $5M \Omega$. Horizontal or vertical P.C. mounting Sub-miniature 0 TW, 5p each. Miniature 0:25W, 7p each

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Draw the planned circuit onto a copper laminate board with the P.C. Pen, allow to dry, the immerse the board in the etchant. On removal the circuit remains in high relief.

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400V: 0.001μF, 0.0015μF, 0.0022μF, 0.0033μF, 0.0047μF, 3p. 0.0068μF, 0.01μF,
0.015μF, 0.022μF, 0.033μF, 3p. 0.047μF, 0.068μF, 0.1μF, 5p. 0.15μF, 6p. 0.22μF,
7½p. 0.33μF, 11p. 0.47μF, 13p.
160V: 0.01μF, 0.015μF, 0.022μF, 0.033μF, 0.047μF, 0.068μF, 3p. 0.1μF, 31p. 0.15μF,
4½p. 0.22μF, 5p. 0.33μF, 6p. 0.47μF, 7½p. 0.68μF, 13p. 1.0μF, 13p.

MULLARD POLYESTER CAPACITORS C280 SERIES 250V P.C. mounting: 0.01μF, 0.015μF, 0.022μF, 3p, 0.033μF, 0.047μF, 0.068μF, 3p, 0.1μF, 4p, 0.15μF, 0.22μF, 5p, 0.33μF, 6ξp, 0.47μF, 8ξp, 0.68μF, 11p, 1.0μF, 13p, 1.5μF, 20p, 2.2μF, 24p,

MYLAR FILM CAPACITORS 100V CERAMIC DISC 0-001μF, 0-002μF, 0-005μF, 0-01μF, 0-02μF, TORS 100pF to 10 ap. 0-04μF, 0-05μF, 0-068μF, 0-1μF, 4p. to 10,000pF,

ELECTROLYTIC CAPACITORS—MULLARD 015/6/7
(µF/V) 1/63, 1.5/63, 2.2/63, 3.3/63, 4.7/63, 6.8/40, 6.8/63, 10/25, 10/63, 15/16, 15/40, 15/63, 2.2/10, 22/52, 22/63, 33/63, 33/16, 33/40, 47/4, 47/10, 47/25, 47/40, 68/63, 68/16, 100/4, 100/10, 100/25, 150/63, 150/16, 220/4, 220/63, 220/16, 330/4, 6p. 47/63, 100/40, 150/25, 220/25, 330/10, 470/63, 7p. 68/63, 150/40, 220/40, 330/16, 1,000/4, 10p. 470/10, 680/63, 11p. 100/63, 150/63, 220/63, 1,000/10, 12p. 470/25, 680/16, 1,000/25, 1,000/16, 1500/10, 2,000/16, 18p. 330/63, 680/40, 1,000/25, 1,500/16, 2,200/10, 3,300/63, 4,700/4, 21p.

SOLID TANTA	ALUM BEAD	CAPACIT	ORS			12p
	μF 35V	2-2µF	35V	22μF 33μF	16V	
	μF 35V	4.7µF		33µF	10	
0.47	μF 35V	6 8µF	25V	47µF	6-3V	
1.0	μF 35V	I0μF	25V	100µF	3∨	

VEROBOARD 2½ × 3½ 2½ × 5 3½ × 3½ 3½ × 3½	0· 24p 28p 28p 32p	0·15 20p 28p 28p	JACK PLUGS AND SOCKETS Standard screened Standard insulated 12p Stereo screened 40p 3-5mm screened 18p Standard socket 20p 2-5mm socket 10p
17" × 2½ 17 × 3½ 17 × 3½ (plain) 17 × 2½ (plain) 2½ × 5 (plain)	85p 120p 76p	67p 108p 52p 41p 12p	Stereo socket 30p 3-5mm socket 11p D.I.N. PLUGS AND SOCKETS 2 pin, 3 pin, 5 pin 180°, 5 pin 240°, 6 pin, 7 pin Plug 12p. Socket 8p. 4 way screened cable, 25p/metre. 6 way screened cable 30p/metre.
2½ × 3½ (plain) Pin insertion to Spot face cutter Pkt. 50 pins		62p 52p 20p	BATTERY ELIMINATOR £1-70 9V mains power supply. Same size as PP9 battery.

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3W CASED AMPLIFIER

3W CASED AMPLIFIER
Polished wooden cabinet 14in × 13in × 9in; containing a sensitive (20µV) 4 valve amplifier with tone and volume controls Gives 3 watts output to the 7in × 4in 30 speaker Also a non-standard tape deck Supplied in good working condition with circuit Standard mains operation, £3-50 (£1). Suitable cassette £1-10 (30p) Spare head 33p. Tape (ex-computer) 75p (20p). Amplifier chassis only, complete and tested (2 × ECC83, EL84, EZ80) ~ speaker £2-50 (40p).

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Contain 2 CV2271 dekatrons, 2 4-digit resettable counters, 2 sealed relays, 2 × ECC81, R's, C s, etc. Only £1-70

COMPUTER PANELS

30b assid t1-10 (30p), 7lb t2-20 (40p), 56lb t12 (t2), 12 high quality panels with IC s, power transistors, trimpots, etc. t2-20 (30p), 100 t13 (t1), Panel with 7 2kΩ trimpots 33p (10p). Pack containing 5014 pin DlL (5s, Inc. 7400, 04, 10, 20, 30, 7474, etc., also MC types t2-20 (20p).

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500 asstd resistors £1-10 (20p); 2,500 £4 (50p), 10,000 £13 (£1); 100,000 £88 + carr 150 poly, ceramic, mica caps \$6p (10p)

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NEW COMPONENTS
400 58 SCR 56p; 200′ 44p; OC71 9p;
14 for t1·10; OC210 8p; 10c t5·50; 741C
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SC40D Triac t1: 34 100 V Rec. 16p;
O5:500V discs; 100 for t1·66; 1 000pf
disc, 100 for t1·55; 80+ 80+ 20µF 350V
10p, 10 for 83p; (30p), 8µF 2·500V t2·20
(50p), 4-poin DN plug 10p, 5-pin 180°
11p, 3 5mm plug 8p, phono 5p 7-pin DN plug 11p, 5-pin DN plug

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EXCLUSIVE: SPECIAL OFFERS

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+ or — Earth with speaker

30, Pair Akai ADM

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BUILD THIS RADIO

BUILD THIS RADIO

Portable MW/LW radio kit using Mullard RF/IF module. Features MW—bandspread for extra selectivity. Slow motion tuning. Fibre glass PVC cabinet. 600MW output. All parts £7.98 (battery 22p), carr. etc. 32p.



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130 Mono control	2 25
unit 3-60	255 Level Indicator
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for II5 4-55	525 120-160mHz VHF
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MW/LW Radio Tuner to use with any amplifier. Features Mullard RF/IF module Ferrite aerial, built in battery. Excellent results. Size $7^{\prime\prime} \times 2_4^{\prime\prime} \times 3_4^{\prime\prime}$. All parts £4-85, carr. 15p.

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Fig. 627

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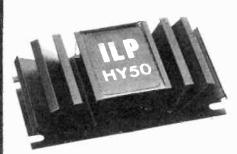
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INPUT IMPEDANCE:
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SIGNAL/NOISE RATIO:
FREQUENCY RESPONSE:
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SIZE:

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±25V. 105 × 50 × 25mm.

Price £5:80 mono £11:60 stereo. Price inclusive of VAT & P. & P.

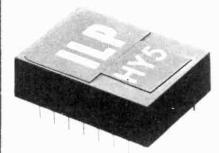
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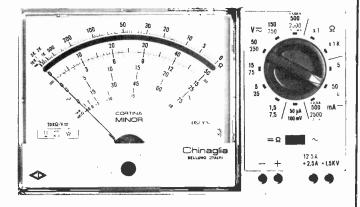
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Radial leads for P.C.B. mounting. Working voltage Radial leads for P.C.B. mounting. Working voitage 240V d.c. 0.01, 0.015, 0.022, 0.033, 0.047, 3p each; 0.068, 0.1, 0.15, 4p each; 0.22, 5p; 0.33, 7p; 0.47, 8p; 0.68, 11p; 1.0, 14p; 1.5, 21p; 2.2, 24p.

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Working voltage 500V d.c.
Values in pF-2-2 to 820 in 32 stages, 6p each;
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2,700, 3,600, 12p each; 4,700, 5,000, 15p each;
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2 circuit with chrome nut and black/white/red/
green or grey unswitched, \$5/\$5/\$5/\$5/\$5/\$8/\$6/\$6/\$17p
Miniature, 3-\$mm, 2-circuit (black), 2 br. cont., \$6/8B

S6/BB

EV CATALOGUE 7

2nd Printing

112 pages, thousands of items, illustrations, diagrams, much useful technical information. The 2nd printing of this catalogue has been updated as much as possible on prices. It costs only 25p post free and includes a refund voucher for 25p for spending when ordering goods list value £5 or more.

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P2	18p
3-circuit, screened, top entry, P3	53 p
3-circuit, screened, side entry, SEP3	55p
Miniature, 3.5mm, 2-cir., screened, P5	130
Maratare, 5 Shini, 2-chi, sercence, 15	
Miniature, 3.5mm, 2-circ., unscreened,	A911002
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and their decades. E24: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

62, 75, 91 and their decades.
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up in units of 100 only.

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P	0.47	_	_	_	_		_	Hp	8р
	1.0	_		—,	_	-	Пp	_	8р
	2.2	-	*****	_		Hp		8р	9p
1	4.7	_	_		Пp	-	8р	9p	9p
	10		_	_	_	8р	9р	8р	8р
	22	_	_	8р	_	9 p	8р	8р	10p
	47	8р	_	9p	8р	8р	8р	10p	13p
	100	9p	8р	8p	8р	9p	10p	12p	19p
IJ	220	8p	8р	9p	10p	10p	lip	17p	28p
П	470	9p	10p	10p	Hp	13p	17p	24p	45p
	1.000	Пp	13p	13p	17p	20p	25p	41p	-
П	2,200	15p	18p	23p	26p	37 p	41p	-	_
	4,700	26p	30p	39p	44p	58p	_	-	_
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Wind the first state (and massiv) 43m 0.040 pins

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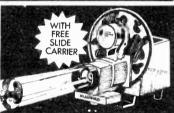
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 $15\frac{1}{2}^{n}\times8^{n}\times4^{n}.$ Complete deck and cover in closed position approx. $15\frac{1}{2}^{n}\times12^{n}\times6^{n}.$ Complete only $\ref{eq:constraint}$

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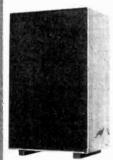
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2	1	71	2-09	0.22
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6	3	70	3.56	0.42
4	4	108	3.96	0.52
10	õ	72	4-67	0.52
12	6	116	5-67	0.52
16	8	17	6-64	0.52
20	10	115	10-23	0.69
30	15	187	13.75	0.97
40	20	232	18-26	1.00
60	30	326	22-52	1-10

00	30	325	EE-95	1.10
30 V	olts	- 17:		
Prim. 2	00-240V.	Sec.	12, 15, 20,	24, 30V.
Amps	Type		Price	Post
0.5	112		£1-58	€0.22
3	79		2-20	0.38
2	3		3.19	0.38
	30		3.96	0.42
4	24		4-68	0.52
ō	51		5-80	0.52
6	117		6.93	0.52
8	88		9-00	0.67
10	м9		10-00	0.67

Prim.	olts 200-240V 9, 25, 33, 4				olts :00-240V. :30, 40, 4	8. 60V.	
Ampa		Price	Post	Amps	Type	Price	Post
0.5	102	£2-11	€0.30	0.5	124	£2·10	£0.38
1 ,	103	3.08	0.38	1 1	126	2.97	0.38
Ι,				2	127	5-77	0.42
2	104	4.29	0.42	3	125	7-15	0.52
- 3	105	5-77	0.52	4	123	9.35	0.67
4	106	7.48	0.52	5	40	11-55	0.67
6	107	11.00	0.67	6	120	13-57	0.82
				1 8	121	16.00	1.00
- 8	118	14.19	0.97	10	122	19-40	1.00
10	119	17-60	0.97	1 12	189	21.62	1.10

	11-00	0 31	. 12	109	21.02	1.10
MINIATURE	AND E	QUIPMEN	T			
Prim. 240V with	screen.					
Volts		Milli	amps	Type	Price	Post
Sec. I	Sec. 2	Sec. 1	Sec.2	No.	2	£
3-0-3		200		238	1.23	0.10
0-6	0-6	500	500	234	1-80	0.10
0-6	0-6	1000	1000	212	1.68	0.22
9-0-9		100		13	1.23	0.10
0-9	0-9	330	330	235	1.48	0.10
0-8-9	0-8-9	500	500	207	2-28	0.25
0-8-9	0-8-9	1000	1000	208	3.03	0.30
15-0-15		40		240	1.23	0.10
0-15	0-15	200	200	236	1.30	0.10
20-0-20	_	30		241	1.23	0.10
0-20	0-20	150	150	237	1.80	0.10
0-15-20	0-15-20	500	500	205	2-97	0.38
0-20	0-20	300	300	214	1.76	0.22
0-20			(No screen)	1116	3.00	0.40
20-12-0-12-20		700 (D/	C) —	221	1-55	0.36
0-15-20	0-15-20	1000	1000	206	3-80	0.38
0-15-27	0-15-27	500	500	203	3.08	0.38
0-15-27	0-15-27	1000	1000	204	3-24	0.38

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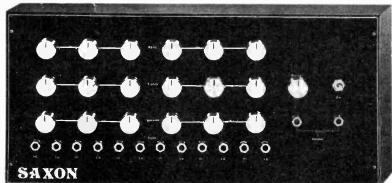
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EMI 341	n. 3 or 8 c	hm C/Mag	. 1.00	Dome Tweeter 8 of Crossovers CN 23 (3		4.
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		with 8in e			. w. A. Ceetti	0
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CM70 Pl	anet stick	metal,		CONDENSER MI	E 600 ohm,	-
	crystal		1-55	uni-dir	-ish D	9-
DM 160 I	ynamic o	omni-dir, b	all 4-95	Cassette Stick Mike Control on/off sw		
	0K/600 ol	hm, uni-di	1.	and 3.5mm J/Ply)	1-
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SOLDER	ING IRO	rs .		spare Blb, etc.)		3.
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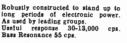


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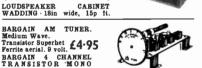
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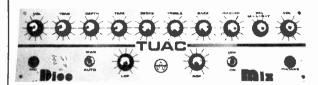
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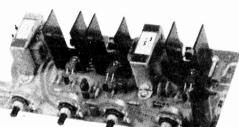
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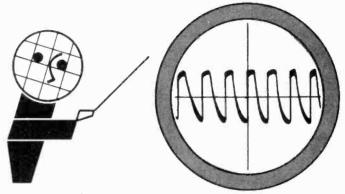
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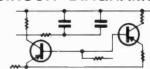
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SOUND POWER

A UDIO power amplifiers find their way into a very extensive and diverse range of equipments and systems, quite apart from the domestic entertainment area. The exuberant folk music of today, whether originated "live" or recreated from discs, is borne along on hundreds of watts. Public Address systems reach out as never before to cajole or inform captive audiences or, just as probably, innocent passers by. Just two of the more blatant uses of sound power.

No one can avoid the fact that amplified sound is very much part of the contemporary scene. The propriety of its widespread use can be argued about but its presence is real

and, presumably, permanent.

Those concerned with technical matters can, at any rate, do their best to ensure that the amplifying medium, for its part, is beyond reproach and does not introduce gratuitously any (further) distortion. In fact, it is generally agreed that this branch of the technology has made impressive progress and audio engineering has reached almost the ultimate in the quest for faultless performance. The evidence lies among the huge variety of amplifier designs which have been developed.

Yet the question could well be posed: are there not too

many designs?

The need for off-the-shelf designs to meet the constantly occurring demands for certain basic units (such as audio power amplifiers and power supply units) which form indispensable sections of many electronic systems, has been a theme of discussion among electronics people over the years. The logic behind this thought is undeniable, but the ideal general purpose design has never materialised, save in certain rather limited areas. So designs for standard routine functions still proliferate.

The continual up-dating of the technology through the introduction of new components is the stumbling block. Electronics is synonymous with change. A well proven standard device, equipment, or system is likely to have a limited life before the inevitable innovation or improvement happens.

Yet even within this volatile field there has perhaps been a rather excessive outpouring of amplifier designs than is

altogether warranted.

There is still a reasonable life expectancy for many designs; and when considering a widely used item such as an a.f. amplifier there is good sense in stabilising the position as long as possible by fully exploiting one single design. If the design is capable of being produced in a number of different "sizes" so catering for the majority of power-requirements commonly called for through the simple expedient of changing component values, the attractiveness of this policy is unquestionable.

This idea materialises in our pages with the introduction of a family of audio power amplifiers. The P.E. Power Slaves comprise a family of four amplifiers having outputs of 20, 40, 65 and 100 watts respectively, suitable for use with hi fi pre-amplifiers. Furthermore any unit can easily be built in a two-channel form with a doubling of output power. Good performance, with low harmonic distortion and good signal to noise ratio, justify the claim that the P.E. Power Slaves are a step towards that goal of a universal power amplification block for embodiment in all kinds of sound reinforcement systems.

F.E.B.

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THE purpose of any audio chain is the accurate reproduction of live or recorded information, be it speech, music or just plain "sound".

Due to the low efficiency of transducers generally, with the possible exception of horn loaded systems, considerable power must be expended in order to produce sounds at an acceptable level; this level will obviously be different for different transducers in different applications. As the basic function is the same, it was felt that, rather than produce a flood of designs for specific purposes, a better approach was to produce a single basic design to fulfil all requirements.

From this article four amplifiers can be built all of them being based on a single circuit configuration. To realise the range of outputs of 20W, 40W, 65W and 100W component changes should be made according to the relevant parts lists.

Although a single amplifier is described the prototype units were designed as double channel, that is, where two amplifiers are contained on a single chassis. This, of course, means a doubling of output capability to 20 + 20W, 40 + 40W, 65 + 65W and 100 + 100W in each case.

DESIGN CRITERIA

Where a low power system is required, there is little to be gained, and much to be lost in producing a power amplifier with an output capability of less than twenty watts into eight ohms, as cost and physical size will not be significantly affected, and performance criteria such as transient handling capability, signal to noise ratio and distortion are not likely to improve.

Turning to the upper end of the power scale, choice of maximum power can be arrived at by examination of the transducers to be driven. Modern composite high power speakers, and individual drive units, will accept continuous sine wave outputs of between 10 and 50 watts r.m.s. Unfortunately audio signals are rarely as predictable as sine waves, but it can be assumed that twice the continuous power can be handled on transients of short duration, thus a power capability of 80 to 100 watts will be required.

When two or more speakers are connected in series, or series-parallel to maintain impedance, then the power requirement can be greatly increased. No allowance has been made for this method of use, as the damping effect of the amplifier is lost, except when units are connected by separate leads to the amplifier terminals, and thus it is felt to be a bad policy.

A further problem arises when two speakers in parallel are driven to a peak level of 100 watts each, that is, a current handling capability of at least 10 amps is required, both in the amplifier and power supply unit, and this is dangerously close to the maximum current of a TO3 transistor (generally 15 amps).

Thus a high quality amplifier should be capable of driving up to 100 watts into a single 8 ohm load, and, to satisfy the requirements of all applications, must have very low harmonic and intermodulation distortion figures, be overload protected, have a high damping factor, and a very high signal to noise ratio.

This final parameter is often neglected, and in most designs a figure of 80dB is considered good. However, a better figure to aim for is 100dB.

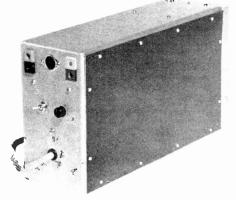
Power amplifiers can be functionally split into two sections, namely a voltage amplifier with low current gain, followed by a current amplifier with unity voltage gain, and at this stage the two sections will be considered separately.

Specification . . .

Power bandwidth Total harmonic distortion Intermodulation distortion Signal/Noise ratio Input sensitivity into 200kΩ Input matched to 600Ω line Overload protected

-6dB from 8Hz to 45kHz 0.02% 0.04% 100dB

1V r.m.s.



amplifier Full bridge output to eliminate "turn-on plop"

May be d.c. coupled throughout for use as a servo

PROVIDING OUTPUTS OF 20, 40, 65, & 100 WATTS FROM ONE COMMON CIRCUIT CONFIGU

VOLTAGE AMPLIFIER

The basic voltage amplifier configuration used for the power amplifiers is shown in Fig. 1. It has the advantage of providing adequate performance without undue complexity and requires no setting up.

The input resistor R1 provides bias current for TR1 and defines the slave amplifier input impedance.

R12, R2 and C2 form an attenuator which determines the proportion of the output fed back to TR2 and hence defines the gain of the amplifier. C2 allows full d.c. feedback to stabilise the output at, or close to 0 volts. Transistors TR1 and TR2 form a differential amplifier, as do TR3 and TR4, the output of this second stage being fed to a common emitter amplifier TR5.

The overall combination will have a very high open loop gain, and will be non-inverting, as a positive going input tends to increase current in TR1 at the expense of TR2, so that TR4 turns on and TR3 off. This reduces the current drawn by TR5 base and gives a positive going output at TR5 collector. Resistors R4, R5, R7, R8 and R9 enable a sensible standing current to be established in each stage, and eliminate any chance of leakage currents unbalancing the amplifier.

The problem of non-linearity in TR5 is overcome by operating this transistor at a constant collector current by bootstrapping the junction of R10 and R11 to the output by means of C3, providing an almost constant current in R11.

COMPOUND TRANSISTORS

The output stage is basically a Class AB compound using quasi-complementary transistors being made up of groups of three as seen in Fig. 3.

These triplet configurations have a number of advantages chief of which are (a) high current gain, (b) the inputs are presented through a single base

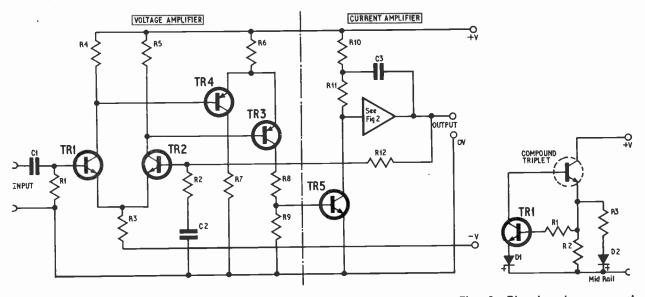


Fig. 1. Basic circuit arrangement for voltage amplifier

Fig. 2. Showing how current limiting is achieved

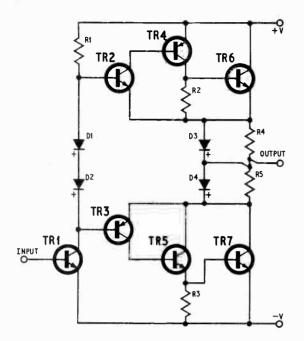


Fig. 3. Compound triplet arrangement for current amplifier

emitter junction which means that the output transistor junction temperature will not affect stability and (c) *npn* output transistors are relatively cheap and readily available.

In the quiescent condition the driver transistor TR1 is biased to provide a mid-rail voltage at the

output. With a signal applied the collector applies alternate drive conditions to the compound triplets so that first one group is conducting and then the other.

A "two stage" resistor is used to link the triplets to the output terminal, consisting of a fairly high value resistor shunted by a power diode and resistor.

The high value of resistance gives good quiescent stability, and the diodes limit the voltage drop under full drive conditions. To ensure reliable operation of the overload protection, a low value resistor is connected in series with each diode, and maximum current can be adjusted by means of this resistor.

OVERLOAD PROTECTION

Overload protection is provided by limiting the output current supplied by either compound triplet. The circuitry that achieves this is shown in Fig. 2 for one triplet only, a complementary circuit being used on the other.

The current supplied to the output terminal generates a voltage across the combination of R3, R2 and D2 which is related to the current flowing. When the current rises to a level sufficient to provide approximately 4.5V, base current is directed through TR1 and D1. Conduction of this transistor means that bias is removed from the triplet so that the output current is limited.

The peak current available before limiting must be higher than the peak current to be supplied to the load, and as this may be 5A, a dissipation of 225W can occur with a short circuit at the output. Because of this, a fuse circuit must be inserted (FSI) to give protection.

The complete circuit configuration for the basic amplifiers is given in Fig. 4.

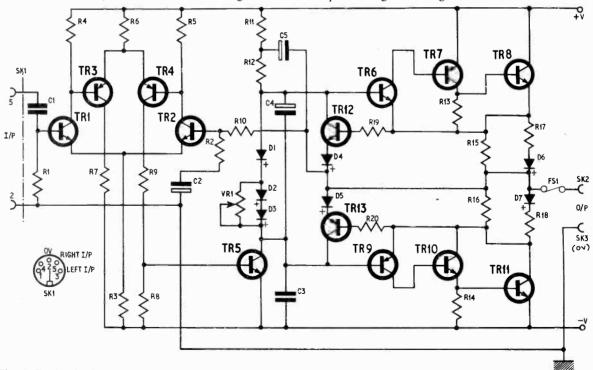


Fig. 4. Basic single channel (right) configuration common to all slave amplifiers. For component value variations, see components lists. The inset shows wiring for a 180 degree, 5 pin socket for two channel. When connecting a remote preamplifier the screen is connected to earth (0V) only at the slave to obviate hum loops

POWER SUPPLIES

Power supplies in a design such as this must provide a fairly steady supply voltage, with a high current capacity when required. As an example, 100 watts into 8 ohms will require a peak current of 5 amps per channel. This implies that for a stereo amplifier with a common power supply a peak requirement of 10 amps is necessary. The mean current drain will be 1/2.8 of this value, or 3.5 amps, and so a transformer rated at 4 amps continuous current would be suitable.

Decoupling should maintain the supply within about two volts of nominal, at full power, for one half cycle of mains current so that:

$$C = \frac{IT}{V}$$
 at peak output

where I is in amps, T in seconds and V volts. The equation then is:

$$C = \frac{10 \times 0.01}{2}$$
 Farads = 50,000 μ F

A composite musical signal can be shown to have a short term mean level at least 12dB below maximum output. Occasional transients will reach maximum output, but these will generally be of short duration, two milliseconds or less being typical.

From similar calculations as above the decoupling requirement for music at -12dB is 5,000 µF and

for a 2mS transient, 6,000µF.

DESIGN CHOICE

If the amplifier is intended for home entertainment (domestic) reservoir capacitors of 6,000 μF are

COMPONENTS . . .

	AMF	LIFIERS			POWER SUPPLIES						
Resistors	20W	_	65W	100W	Resistors	20W	40 W	65 W	100W		
R1				200kΩ	R1	12k Ω	15k Ω	18k Ω	20k Ω		
R2, R6	39kΩ		68kΩ	86k Ω	R2	100Ω	100Ω	100Ω	100Ω		
R3, R4, R5	430kΩ	$620k\Omega$	750kΩ	910kΩ	R3, R4	200Ω	300Ω	360Ω	470Ω		
R7	$91k\Omega$	120k Ω	150k Ω	160k Ω	R5	300Ω	300Ω	300Ω	300Ω		
R8	6·2kΩ		6·2kΩ	6·2kΩ	R6	10Ω	10Ω	10Ω	10Ω-		
R9	82k Ω	110k Ω	130k Ω	150k Ω	R7 see text						
R10	200 k Ω		200kΩ	$200k\Omega$	R8	10k Ω	12k Ω	15k Ω	18kΩ		
R11, R12	15k Ω	$20k\Omega$	24kΩ	$27k\Omega$	R9	1k Ω	1kΩ	1kΩ	1kΩ		
R13, R14	100Ω	100Ω	100Ω	100Ω	R10, R11	$2k\Omega$ (1W)	$2k\Omega$ (1W)	$2k\Omega$ (1W)	2kΩ (1W)		
R15, R16	10Ω (1W)	10Ω (1W)		10Ω (1W)	Potentiometer	s					
R17, R18	0·11Ω	Ω 80.0	0.07Ω	0.06Ω	VR1	1kΩ	$1k\Omega$	$1k\Omega$	1kΩ		
0.03	Ω/cm , 26 s	.w.g. resis	tance wire		****		all carbon	lin e ar			
R19, R20	4·7kΩ	4·7kΩ	4.7kΩ	4·7kΩ	Ci4						
All ½W 5% m	etal oxide (except whe	ere otherwi	se stated	Capacitors	10,000μF	10,000μF	10,000μF	10,000μF		
					C1, C2	(40V)	(50V)	(63V)	(63V)		
					C4, C5	10.000μF	10.000μF	10,000μF	10,000μF		
Capacitors	0·1μF	0·1μF	0·1µF	0·1µF	C4, C3	(40V)	(50V)	(63V)	(100V)		
C1	$0^{\circ}1\mu$ 1	Ali 30V ce		٠ ١٣٠		(401)	All electro		(1001)		
C2	10μF	10μF	10μ F	10μF	C3	100pF	100pF	100pF	100pF		
C2	ιομι	All 16V ta				,006.	. оор.	, , , ,			
C3	100pF	100pF	100pF	100pF	Transistors	D.C.004	D.C004	BC204	BC204		
00	.006.	All silver	mica	•	TR1	BC204	BC204				
C4	22μF	22μF	22μ F	$22\mu F$	TR2, TR3	MPS-001	MPS-UUI	MPS-U01 MPS-U51	MDS LIE1		
Ŭ,		All 16V ta	ntalum	•	TR4 TR5	2N3055		SDT9203			
C5	1μF	1μF	1μF	1μF		2143055	3D 19203	3D 19203	3D 19200		
	•	Áll 63V p	olyester		Rectifiers						
					D1–D4	REC46 (150V, 10A)					
D . 4 4' 4					D5		BZY88—C	C6V8			
Potentiometer VR1	rs 2·2kΩ	2·2kΩ	2·2kΩ	$2\cdot 2k\Omega$	Transformer						
VKI	Z. ZK 22	Z.ZV27	2 2 1 3 2	2 2132	T1	24-0-24 V	30-0-30V	33-0-33V	36-0-36V		
						@ 4A	@ 4A	@ 5A	@ 6A		
Diodes					Cha	naes for	Standard	Supplies			
D1-D1\$	1N914	1N914	1N914	1N914				g version			
DS-D	1 N5401	1Ñ5401	1 N5401	1N5401	Capacitors	come ama		.g			
					Gupuonoro	20W	40W	65 W	100W		
1					C1, C2	6.000µF	6,000µF	6,000µF	6,000µF		
Transistors	0.010.40.4	2N2484	2N2484	2N2484	(Domestic)		(35V)	(40V)	(50V)		
TR1, TR2	2N2484 BC212	2112404 MDS_I 51	MPS-L51		C1, C2	$10,000 \mu F$	10,000μF	10,000μF	10,000μF		
TR3, TR4	ME1120	ME1120	ME1120	ME1120	(Monitoring)	(25V)	(35V)	(40V)	(50V)		
TR5 TR6	BC182		MPS-L01								
TR7	MPS-1157	MPS-U57	MPS-U51	MPS-U57	Rectifiers		DEC 40 A	(100)/ 44)			
TR8, TR11	2N3055	2N3055	SDT9203	SDT9203	D1-D4		REC 43A	(100V, 4A)			
TR9	BC212	MPSL51	MPSL51	MPSL51	Transformer						
TR10	MPS-U0	MPS-U07	MPS-UO	MPS-U0	T1 18-0-18	V 22-0-	22V 25	-0-25V	30-0 - 30V		
TR12	BC184	BC184	BC184	BC184	@ 3A	@ 4/		4A	@ 4A		
TR13	BC214	BC214	BC214	BC214	G 47.	9	-	ouglas	Douglas		
								T106	MT123		
							141		, 120		
Miscellaneous					Miscellaneous			1.04	-1		
FS1—	1A	1A	1A	1A	S1—Double	pole ma	ins on/off	, LP1—m	ains neon		
Heatsink	60D-0600	-A-1 (Mars	ston) (100V	V and 65W)	FS1—IA						
							A STORE SHOWING COMMENTS	Market State of the State of th			

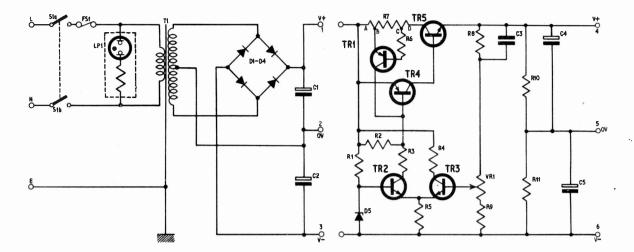


Fig. 5. Circuit for p.s.u. For the domestic version of amplifier the power rails are pins 1, 2 and 3. The monitoring version takes its supply from pins 4, 5 and 6. The current sensor R7 is made up from 26 s.w.g. Constantan resistance wire (0·03 ohm/cm). The tapping resistances are as follows: A to B \leq 0·02 ohms; B to C = 0.07 ohms; C to D \leq 0.02 ohms. Supply voltages for the 20, 40, 65 and 100W versions are approximately \pm 25V, \pm 3eV and \pm 4eV respectively

adequate assuming that neither transients nor very loud passages coincide exactly in time if two channels are being used.

For "monitoring" quality, a higher value capacity, possibly 10 or 15,000µF should be used, or alternatively a stabilised supply particularly when used for stereo to reduce cross talk.

A composite circuit diagram for the domestic and monitoring versions of the p.s.u. is shown in Fig. 5.

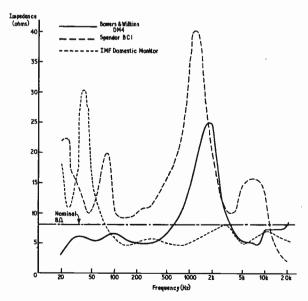


Fig. 6. Showing how speaker impedance changes with frequency

For the former points 1, 2 and 3 are connected to the amplifier. For the latter points 1 and 2 connect to the stabiliser circuit; no connection is made to 2.

The output mid-point in this case is derived using resistor and capacitors.

OPTIMUM LOAD

Optimum load impedance for all versions is 8 ohms but any load can be accommodated at reduced power, bearing in mind the peak voltage and current limitations.

Fig. 6 shows typical impedance curves for high quality domestic speakers demonstrating that the impedance is anything but constant. Indeed, it is not unusual to find dips of impedance as low as 1.5 or 2 ohms, causing severe current overload at critical frequencies.

IMPORTANT

Two BA148 diode suppressors MUST be added to Fig. 4 as follows—Cathode and anode to \div V and SK2 and SK2 and — V respectively. In Components List for all versions TR7 and TR10 should be MPS-U57 and MPS-U07; for the p.s.u. TR2, TR3 should be MPS-U07 and TR4, MPS-U57. (See next month.)

Next month constructional details for the amplifiers will be given

THE SUN, BY SKYLAB

From the last of the experiments aboard Skylab has come new knowledge and exciting data. There is no doubt that a new era has opened for astronomers and as a result of what has been noted many programmes will be instituted in solar physics.

Much of the success is due to the (ATM) Apollo Telescope Mounting. It enabled a new view of the solar system, the galaxy and, most

important, the Sun itself.

Free of the atmospheric absorption the telescope showed that the Sun, far from being "quiet" at the near sunspot minimum was extremely active. This alone has justified the mission and also confirmed the need for man in space.

Practical Electronics received an invitation from the Institution of Radio and Electronic Engineers to attend the Clerk-Maxwell Lecture, "Man in Space" by Werhner von Braun. More than twice the number of persons attended than the seats available. Before the lecture began the writer had a conversation with Werhner von Braun and certain definite conclusions regarding the progress of physics emerged. In his words "... the time has come for another Maxwell to reorder our thoughts and set in their proper place the ever increasing number of particles, the black holes and the quasars." He also gave some information based on the early examination of some of the data.

The medical evidence is now clear and it is accepted that man could live in space indefinitely. The adaptability is not confined to man for the spider which was taken aboard Skylab was able to spin a perfect web. The pictures of the first attempts were not only comic but extremely interesting. It needed only a short while of accommodation before things were under control. The peculiar zigzag of the circles and the cross ties soon assumed the familiar shape and life went on.

The astronauts also attempted to move round the walls at a high rate to set up artificial gravity. In this they were not successful.

GREAT BALLS OF FIRE

One of the most spectacular events was a pulsing stream of plasma from the Sun which had a regular rhythm. More than forty of such events were recorded. An unexpected bonus was the ability to observe with the coronagraph these great blobs, as large as the



Sun itself, as they moved up through the corona at a speed of 400km per second.

The variation of the intensity of the corona over the two hemispheres of the Sun were not anticipated. One section of the Sun seemed to be in the nature of a hole at a lower temperature than the rest of the surface. This was borne out by the X-ray pictures. It also appears that the corona is composed of closed loops which fit the magnetic line structure. In both the X-ray measurements and the white light coronagraph pictures this appears.

There were a number of instances, where the conditions for sunspot appearance appeared outside the normally observed belts. Some of the "holes" indicate the source of disturbances of the solar wind which effect terrestrial conditions.

There have been tens of thousands of frames taken of the coronal conditions, more than has been observed in a thousand years of eclipses. Continuous observations with the facility to select and also combine different types of telescope, made the *Skylab* control console a sort of astronomical organ on which the melody of the heavens could be produced. Nothing like this success was expected and now the knowledge of what can be done in this field must ensure that the next decade will add an unprecedented amount of data for study.

More than 600 pictures a day were recorded and the number of bits in each picture amounted to 108. Among the many and continuous hours of study covering everything in the vicinity of the Sun there were two solar eclipses,

many eruptions and flares on the Sun itself, the perihelion of Kohoutek and a transit of Mercury.

The combination of control from the Johnson Spaceflight Centre and astronauts aboard Skylab resulted in a magnificent scientific achievement. It is worth noting that the planned tasks for the crew were restricted to the pointing of the telescopes, target control and to collect and replace cameras and film. But, in fact, the crew did much more than that and it was their resource that made possible the full speed running of the observations. Despite some early difficulties, which were overcome, the crew settled down to do a job for which posterity must ever be grateful.

The cost of this mission was high and many thought wasteful. None would say this now and indeed, there is in this very extensive period of man in space, the certainty that money for the future missions will be forthcoming.

MORE OF VENUS

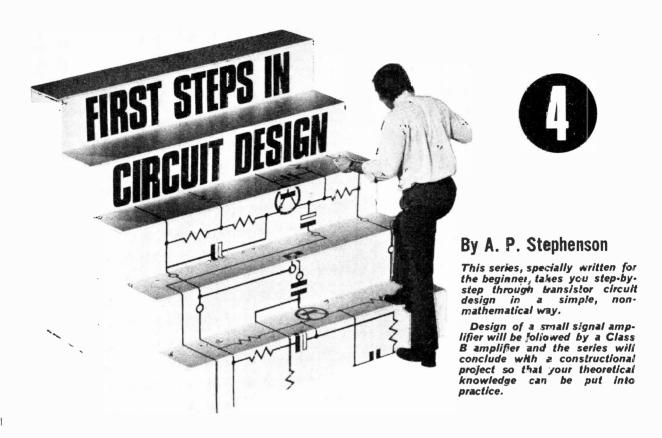
The Venus fly-by of Mariner 10 has been a successful mission. But some of the data obtained will need months to extract the information sought.

The spacecraft approached Venus from the dark side and the first pictures showed the cusps which appear extended far beyond the normal. This is thought to be due to the atmospheric refraction. The scattering of light beyond the terminator will help to an understanding of the upper atmosphere. Haze shells were observed and these will reveal more data after computer processing.

One important aspect is the layering observed in the atmosphere. This is not yet understood nor is the material of the cloud strata.

There was great penetration of the atmosphere by the two radio probe signals. Though physical occultation began at about 10.17a.m. radio signals were still being received up to 17 minutes later. The aerial on Mariner 10 was designed to change its pointing direction in order to compensate for the bending of the radio waves which takes place when travelling through the Venus atmosphere.

Much is expected of this technique in the observation of the density of the atmosphere and it is hoped also to resolve the differences of temperature given by earlier Mariner missions and by the USSR spacecraft Venera. Helium is known to be a constituent but it would have to be at least 3 percent to account for the differences of temperature found.



So FAR we have only looked at the voltage divider method of biasing a transistor stage. This month we will look at another method which is simple and economical. These advantages are, however, at the expense of other parameters.

Once we are satisfied with a single stage amplifier, the

problem then arises of how to connect this into the system so that it does not disturb the driving or driven circuits. Sometimes direct coupling is possible but often we must make use of a d.c. blocking capacitor. The value of this component must be such as to maintain the required bandwidth of the system.

4.1. ALTERNATIVE BIASING ARRANGEMENT

There is an alternative to the R1, R2 divider (Part 3) known as "collector to base feedback bias". Only one resistor is needed instead of two and some bias stability is possible even without the inclusion of $R_{\rm E}$ (see Fig. 4.1).

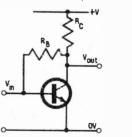
The resistor R_B provides bias and also some negative feedback which tends to stabilise the collector current

against changes in h_{FE} .

Suppose that h_{FE} is higher than predicted: this would tend to increase I_C which in turn would cause the output voltage to fall. This fall would be passed via R_E to the base causing a decrease in I_C . The base bias conditions are therefore stabilised.

HIDDEN FEATURES

Although the method appears delightfully simple and economic, there are certain hidden features which the designer must understand.



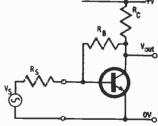


Fig. 4.1 Simple circuit illustrating collector to base feedback (left)

Fig. 4.2 The source resistance can be seen to be an important parameter in the calculation of the circuit gain (right)

(a) The feedback is in parallel with the input signal and therefore requires extra current drive from the signal input to the stage. In other words, the stage input resistance $R_{\rm IN}$ is lowered. In fact, the input resistance consists of two resistors in parallel, one of them being the normal $r_{\rm IN} = h_{\rm te} \times r_{\rm e}$ and the other equal to $R_{\rm B}$ divided by the gain A. This is easily understood when it is realised that the signal voltage is at one end of the $R_{\rm B}$ and the output voltage (A times the input signal) is at the other.

Thus A times as much current must be provided by the signal, which is equivalent to saying that R_B behaves as if it were A+1 times smaller and across

the signal input.

(b) There is no actual reduction in voltage gain, as far as the gain from base-in to collector-out is concerned. The voltage gain is still

$$A = R_{\rm C}/r_{\rm e}$$
 or $R_{\rm C}/(r_{\rm e} + R_{\rm E})$ (If $R_{\rm E}$ is present).

However the lower input resistance has the effect of lowering the gain measured from signal e.m.f. to collector output. The source resistance of the signal (R_S) in conjunction with R_B forms a kind of operational amplifier configuration (see Fig. 4.2).

If A was very high, the total gain from V_s to V_{out} would be approximately

$$A' = R_{\rm B}/R_{\rm S}$$

where A' is defined as the gain with feedback. Thus the gain is independent of the transistor.

This, however, is a gross oversimplification of the situation because the transistor gain A in such a simple single stage amplifier would not be "very high" and $A' = R_B/R_S$ would be greatly incorrect.

This analogy with the operational amplifier was only given as an aid to comprehension. It is always good practice to view circuit ideas from widely differing angles.

4.2. AN ALTERNATIVE EQUATION FOR GAIN

The previous circuit had no emitter resistor $R_{\rm E}$ which was the reason why the gain A was so high. Although a high gain may often be a desirable feature, there are two main reasons against achieving it by the omission of $R_{\rm E}$.

- (a) The gain is not too predictable because of its absolute dependency on r_e . The equation $r_e = 25/I_c(mA)$ is very useful as a rough guide but it is well to remember that the equation is only approximate. The inclusion of R_E in the gain formula, although lowering the gain increases the accuracy of the equation (particularly if $R_E \gg r_e$).
- (b) A more important reason is the rather remarkable twist of the gain equation when the bias is set to allow the collector to rest at half the supply voltage.

If $V_{\text{out}} = \frac{1}{2}V_{\text{ce}}$ then $R_{\text{C}} = \frac{1}{2}V_{\text{ce}}/I_{\text{C}}$. Now $A = R_{\text{C}}/r_{\text{e}} = \frac{\frac{1}{2}V_{\text{cc}}}{I_{\text{C}}} - \frac{25}{I_{\text{C}}} = V_{\text{cc}}/50$.

But remembering that the figure "25" is millivolts, we must multiply by 1,000, giving

$$A = \frac{V_{ec} \times 1,000}{50}$$
, thus $A = 20V_{ec}$

This means that the gain of any grounded emitter stage operating without an emitter resistor is simply 20 times the supply voltage (assuming the collector is at half the supply voltage).

This applies whatever collector current we use, so it is impossible to set the gain at any other value than $20V_{ce}$ (unless of course we use negative feedback).

It is easy to see why $R_{\rm E}$ is almost always present in grounded emitter amplifiers.

4.3. GOUPLING CAPACITORS

It is often required to connect two stages together by allowing the output signal of one stage to become the input signal of the next.

Sometimes the coupling can be achieved without disturbing the d.c. bias conditions, but this is sometimes

unavoidable.

This problem can be overcome with the use of a "coupling capacitor" which enables the signal to pass relatively easily but blocks any d.c. from affecting the next stage.

CAPACITOR VALUE

The question is how big must the capacitor be. The answer depends on two pieces of information—the lowest frequency that the signal is likely to be, and the input resistance of the stage which the capacitor is to feed.

The skeletal circuit is shown in Fig. 4.3.

The capacitive reactance $(X_{\rm C})$ and $R_{\rm IN}$ form a voltage divider, so it is clear that $X_{\rm C}$ should be considerably less than $R_{\rm IN}$ at the lowest frequency which it is desired to pass.

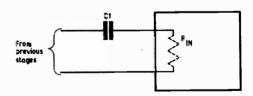


Fig. 4.3 The calculation of the coupling capacitor value is determined by the input resistance R_{IN} of the following stage

The simplest way is to find the value of C which makes $X_C = R_{\rm IN}$ and then use the next highest preferred value.

Suppose the lowest frequency to be passed is 100Hz and $R_{\rm IN} = 1 \, \text{k} \, \Omega$.

Then if $X_C = R_{IN}$, then $1/2\pi fC = R_{IN}$ which gives $C = 1/2\pi fR_{IN}$.

Inserting values gives $C = 1.59\mu\text{F}$ so the next highest preferred value, $2\mu\text{F}$ would be used.

RULE OF THUMB

A neat little rule of thumb to save a lot of fiddling with powers of ten is

 $C = 0.2/f \times R_{\rm IN}$

where f is in kHz, $R_{\rm IN}$ in kilohms and C in μ F. The value for C is in no way critical, providing it is at least the value calculated by the method above. Twice or even ten times larger will not matter apart from the cost.

4.4. USE OF THE EMITTER FOLLOWER STAGE

An emitter follower stage provides an easy solution to the problem of matching from a high resistance signal source to a low resistance load.

Unlike the normal grounded emitter stage, the emitter follower has no voltage gain, in fact, there is a slight voltage loss. There is, however, a substantial current gain as a signal is passed through, because of the apparent change in source resistance.

Treating the emitter follower as a black box, the effect on the input signal is as shown in Fig. 4.4.

The figures shown in the figure are purely arbitrary and are chosen only to illustrate how the source

> EMITTER. FOLLOWER

Fig. 4.4 An emitter follower transforms a high input impedance into a low output impedance

resistance can be substantially lowered by passing through the stage.

CURRENT GAIN

The concept of current gain can be appreciated by calculating the two short circuit currents.

If the original circuit was shorted out the current would be

 $i = 1 \text{V}/100 \text{k} \Omega = 0.01 \text{mA}$

If the output was shorted, the current would be

 $i = 1V/1k\Omega = 1mA$

This shows that a theoretical current gain of 100:1 has been achieved. (The short circuit current dodge is used on paper only—don't take it literally and start sticking screwdrivers across the terminals.)

It must not be supposed that an actual emitter follower would provide a magical solution. The successful design of such a stage requires quite a bit of fiddling with values to get the input resistance as high as

For example, Fig. 4.4 presupposes that the input resistance of the emitter follower stage is much higher than the $100k\Omega$ signal resistance. Unless this is so, it is clear that the signal would be attenuated by voltage divider action between the $100k\Omega$ and $R_{\rm IN}$ of the emitter follower.

4.5. TYPICAL CIRCUIT AND EQUATIONS OF THE EMITTER FOLLOWER

An emitter follower circuit using voltage divider feed for base bias is shown in Fig. 4.5.

We are not concerned with the design factors for setting up the correct d.c. conditions, this problem is dealt with elsewhere. Instead, we state some important equations which decide the voltage gain, input resistance and output resistance.

1. VOLTAGE GAIN

For most practical purposes, the voltage gain is nearly unity, though it is as well to know the following formula in case some badly chosen values for RE lower the gain to a small fraction.

Voltage gain from V_{in} to $V_{out} = R_E/(r_e + R_E)$. Since re is the internal emitter resistance and will seldom be in excess of 100Ω or so, it it easily seen that the gain will be about 1, providing RE is in excess of 1kΩ.

2. INPUT RESISTANCE (RIN)

This is the same basic formula which has already appeared in connection with the grounded emitter stage $R_{1N} = R1$, R2 and r_{1n} in parallel

(remember that $r_{in} = h_{te}(r_c + R_E)$).

3. OUTPUT RESISTANCE (RCUT)

This is a bit complicated unless dealt with in two

 $R_{\text{OUT}} = R_{\text{s}}/h_{\text{fe}}$ in parallel with R_{E}

where $R_s = R1$, R2 and r_s in parallel.

If this is too wearisome to calculate and a very rough equation is all that is needed, then use

 $R_{\text{OUT}} = r_{\text{s}}/h_{\text{fe}}$.

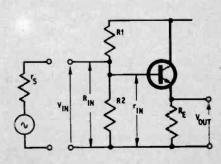


Fig. 4.5 Typical emitter follower with voltage divider for base bias

4.6. D.C. COUPLING

Two stages may be coupled together directly only if the d.c. voltage output of the first stage is compatible with the d.c. bias requirements of the second stage.

A typical example which serves to Illustrate the system is the circuit of Fig. 4.6 which shows a normal grounded emitter amplifier feeding an emitter follower stage.

Transistor TRI provides the voltage gain and TR2 provides a low resistance (high current) output.

The gain of the first stage = $R_{\rm C}/r_{\rm e} + R_{\rm E} = (84 {\rm k}\Omega/{\rm km})$ $1k\Omega + 250\Omega) = 67.$

The second stage contributes no further voltage gain.

STAGE INPUT RESISTANCE

The stage input resistance is r_{in} , RI and R2 in parallel, although it is reasonable to ignore R1

 $r_{\rm in} = h_{\rm fe}(r_{\rm e} + R_{\rm E}) = 100(250\Omega + 1 \,\mathrm{k}\,\Omega) =$ 125kΩ, which, in parallel with R1 gives a value for $R_{\rm IN}$ of about $45 k \Omega$.

Currents through TR2 would be determined first. 1mA is chosen for collector current, so assuming a pessimistic h_{FE} of 100, the base current of TR2 would be 10uA.

The divider feed for the bias of TR2 is the chain $R_{\rm C}$. TR1 and $R_{\rm E}$ of the first stage which is passing 0-1mA (which is ten times the base current of TR2).

Again taking an hrE of 100 for TR1, the base current would be 1µA, which is fed by a divider chain taking 10µA.

The "rule of ten" has thus been used throughout.

OUTPUT RESISTANCE

The output resistance of the emitter follower is a bit tricky. The equation, as stated previously, is Rour = R_S/h_{le} in parallel with R_E . The term R_S is the source resistance of the signal feeding the base, which, in this case, is the output resistance of the first stage.

To a rough approximation this may be taken as the collector resistance, $R_{\rm C}$, which is 84k Ω . Thus $R_{\rm OUT}$: $84k\Omega/100 = 840\Omega$. (The $9k\Omega$ R_E which strictly is in parallel may be disregarded.)

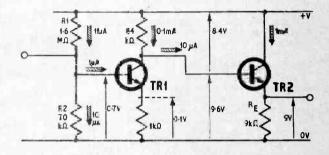


Fig. 4.6 A two stage d.c. coupled amplifier showing typical component values

4.7. D.C. FEEDBACK LOOPS

Leaving the emitter resistor unbypassed has so far been the only example of negative feedback. A more obvious example is frequently used in direct-coupled stages and, to illustrate such a system, the circuit previously discussed will be modified to include a d.c. feedback loop (see Fig. 4.7).

The circuit, including the voltages and currents are exactly the same, apart from the method of biasing the

Instead of the customary divider across the supply rail, the bias feed is taken from across the output resistor.

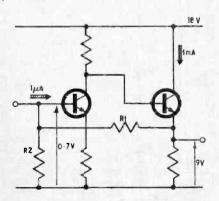


Fig. 4.7 Negative feedback stabilises the output voltage but reduces gain

This is not such a violent change as appears at first sight. After all, there are nine volts available so why not use them and obtain the added advantage of a d.c. feedback loop?

In choosing the values for R1 and R2 we must be careful not to disturb the output circuit too much. It is drawing 1mA so it will hardly be aware of the theft

if we divert say $10\mu A$ from R_E to feed a divider. Since R2 must drop 0.7V, R1 must drop the remaining 8.3V which means that if $10\mu A$ is flowing the values of R1 and R2 are calculated as follows:

$$R1 = 8.3V/10\mu A = 830k\Omega$$

 $R2 = 0.7V/10\mu A = 70k\Omega$

STABILITY

Such a circuit has very good stability in spite of wide tolerances in the resistors RI and R2. In fact, it is astonishing how the output appears to lock at 9V

Suppose that the output tends to drift downwards: this will pull the base of TR1 down which causes the collector of TR1 and base of TR2 to rise. The output tendency is therefore to rise which, due to the feedback loop is providing a correcting influence on the tendency for the input to fall.

The original gain before feedback is, of course, reduced from its previous value. This is one of the penalties to be paid for the benefits c! negative feedback.

Continued next month

THERE must be many PE readers who would never call themselves musicians but who have an interest in music, be it as listeners or as vamping-left-hand pianists. Some of them are no doubt in the throes of building the units which go to make up Mr. Shaw's Synthesiser and are wondering what on earth to do with it when it's completed. Others may not want to begin any musical project because they feel that their creative or recreative talent is nil.

Yet others may be musicians like myself who have a smattering of electronic theory and practice up their sleeves but are afraid to launch any large-scale project because they fear they may never get the finished article to function. I hope this feature may allay the fears of all these people and, indeed, offer some constructive help.

Facilities In the U.K.

Until quite recently the musician, particularly the composer and/or arranger has had to sit on the periphery of technological progress, using a small amount of electronic equipment in a limited kind of way, unable, through lack of know-how, to extend his innate creativity.

In this country it is not just knowhow that has been lacking, but the facilities needed to acquire expertise. Although there has been a tremendous surge of interest in electronically produced or processed sound over the past decade or so we still have not managed to produce a chain of national studios where the musician can work with first-rate equipment and expert technical assistance.

It is true that we have the BBC Radiophonic Workshop, well-equipped and staffed by experts, but, alas, Auntie Beeb tends to hog the show by restricting its output to broadcast material in the shape of incidental music and sound effects and a few (all too rare) commissions from composers outside the Corporation.

Numerous private studios exist too — those of Tristram Cary, Daphne Oram and Peter Zinovieff immediately spring to mind, along with such institutions as the University of York and Goldsmith's College, London—yet most of us cannot hope to experience this kind of Utopian dream. Those of us who live and work away from the metropolis are, as in many things, particularly poorly off.

A studio at home

However, having said that, there is not too much cause for gloom. Magazines like PE have made it abundantly clear that the imaginative person can build quite a sizeable and versatile studio at home, quite apart from the commercial market.



Over the past few years PE has included in its pages features dealing with some of the electronic equipment available to the musician. More frequently there have appeared constructional projects specifically geared to the amateur musician: Waa Waa, Fuzz, Phase, and Sound Bender effects units; and complete instruments such as the Electronic Organ, Electronic Piano, and the Voltage Controlled Synthesiser.

The time is ripe, then, for the musician to join in and help bring the electronics engineer or experimenter and the creative artist together. As a professional musician whose creative work calls increasingly for more and more electronic equipment, I welcome the appearance of this page as a meeting place for the exchange of information, ideas, and ventilating of nagging questions.

Until projects such as the PE Synthesiser appeared (as far as I know the first design to be published for the home constructor to build from scratch) most of us tended to sink further into the slough of despond in the belief that only a fabulous win on the pools could get us reasonably well-equipped.

By beginning in a modest way, building our equipment module by module, we can now keep pace with our pockets whilst steadily increasing our versatility in the creative music sense.

Getting started

I am often asked, particularly by secondary school music teachers who (quite rightly) feel that there is an honourable place for electronic music in the curriculum, what is necessary by way of basic equipment in order to get started.

Assuming that "canned" rather that "live" electronic music is envisaged, then I would suggest a decent half-track stereo tape machine with three or more speeds, sound-on-sound facilities and a couple (maybe only one at first) of reasonable quality microphones; this should keep you going for quite some time.

You may be wondering why I recommend a half-track tape machine when the present-day quality of quarter-track machines is very high indeed. It's not a matter of quality in this instance. Amongst the tricks used in the manipulation of recorded sound one of the simplest is reverse playback, and on i а standard quarter-track machine, stereo or mono, or a halftrack mono machine, reversal of the tape direction merely gives playback of the recorded material on the new channel(s) in a forward direction. On a half-track stereo machine turning the tape over reverses the material on both channels, incidentally reversing the stereo direction too.

Tape manipulation

Strictly speaking, using only microphones, you are not going to produce electronic music, since the microphone will only pick up air-borne sounds, most of which may well be "natural" in origin—dried peas in a can, hand claps, vocal sounds, or the jangle of keys, and maybe acoustic musical instruments. But by varying the tape speeds in recording, superimposition and reverse playback an amazing variety of sounds can be achieved. For the price of an editing block and a reel of splicing tape the sounds can be linked together to form a satisfactory sequence.

Anyone who has persevered with this kind of tape manipulation—and it is, believe me, a time-consuming activity—will tell you that the results can be miraculously electronic in effect.

In case you think this approach to creative sound a bit limiting, then I suggest you go out and buy a recording of "Variations for a door and a sigh" by Pierre Henry (Philips 4 FE 8504). Henry is one of the leading lights in the French movement which began in the early postwar years and produces so-called "musique concrète".

What can be more limiting than the creakings of an old barn door? Yet Pierre Henry manages, through careful handling of the granary door and a bit of superimposition back at the studio, to produce an incredibly wide range of sounds which are as musical as they are amusing.

There is a little paperback, too, by Terence Dwyer entitled "Composing with Tape Recorders" which offers invaluable advice.

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BFS 95	40	70 MHz	17p	2N 5179	20	900 MHz	40p									
BFX 12	25	210 MHz	10p	GERMANI				100000	The second second	NORTH WOMEN	NESSEENE DAMAGE	NAME OF TAXABLE PARTY.	ONLY OF THE PERSON	R. H. S.	THE RESIDENCE OF THE PARTY OF T	NAME OF TAXABLE PARTY.
2N 2906	60	200 MHz	15p	ACY 44	50	1 MHz	10p	\$555K	Contract of				The second second			
2N 3702	40	100 MHz	11p	ADY 26	80	75 watta	£1	4								-
2N 3703	50	100 MHz	12p	AF 124	20	75 MHz	20n	1990				7			148	25
BILICON P				AFY 19	32	350 MHz	20p 20p	East	No. of Concession, Name of Street, or other party of the Concession, Name of Street, or other pa						Harry Thomas St.	
BC 106	30	150 MHz	15p	ASY 32	25	5 MHz	10p				THE RESERVE OF THE PARTY OF THE	AND DESCRIPTION OF THE PARTY OF	ACTION IN COLUMN TO THE OWNER.			
BC 108	30	150 MHz	150	ASZ 21	15	450 MHz	20p	ŀ								
BF 179	225	125 MHz	40p	GET 113	32		10p	linve	rter	Tra	nsforme	rs	13/15	watt	(circuit	•
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BFW 58	60	80 MHz	150	OC 123	50	1 MHz	10p		inc	clude	ed)					70p
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BFX 86	40	50 MHz	17p	2N 1307	30	10 MHz	15p	1.00	rront	Foc	onomy''	Tran	neietoi	· (600r	m Δ \	50p
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2N 718	60	60 MHz	12p	BFR 64		1,200 MHz	E1	D		10-		4				45-
2N 753	25	250 MHz	12p	BLY 89A	35	650 MHz	E5 E5	Hes	istors		pacitors	to s	iuit			15p
2N 744	20	300 MHz	12p	BLY 93A	60	500 MHz	£5	l .			•					
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2N 2220	60	250 MHz	15p	2N 709	15	800 MHz	15p									
	VE DEVICE	E8		-		· 4 GHz	E40	٠,	ong I		,				(pai	r) 30p
CL 8300	Gu	nn effect pecil	lator			-5 GHz	£10	1 (8	hort	lead	1)				(pai	r) 20p
CL 8370		ditto				-5 GHz	£10	٠,			,				(Pa.	,
CL 8380		ditto				· 5 GHz	E10	W/hi	te En	ame	elled Cas	:e 1	8 or 2	1in		70p
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last copper strips on the board are joined by means of a link wire which also connects to one or more intermediate strips. This forms a link between the output and one or more of the inputs according to how the board is wired.

By removing the board it can be turned over and plugged into the same connector making a completely different link or patch between the inputs and outputs. The opposite end of the board can also be used in this way making a total of four different connections available from each board (see Fig. 2).

The programme cards can be numbered for their particular function, e.g. board No. 5 will connect output 3 to input 7, etc.

J. A. Knife, Dagenham, Essex.

PATCHBOARDS FOR SYNTHESISER

Synthesiser patchboards consist of a matrix of jack sockets by which any combination of input and output signals may be connected by inserting a shorted jack plug into the appropriate socket. A patch board for a medium sized synthesiser would require about 14 inputs and 20 outputs, making a total of 280 jack sockets which at 10p each plus the cost of plugs adds up to around £30.

The system to be described has two advantages: a system as previously described would cost only £6, and this type of patchboard can be used to pro-

gramme the synthesiser.

The patchboard is made from 20 sixteen-way edge connectors, the 0.15in matrix size being the most practical. Each connector has its first and last pins wired to the output (Fig. 1). The remainder are wired to separate inputs.

In practice a piece of Veroboard also of 0.15in matrix is plugged into the connector. The first and

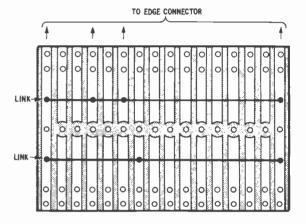


Fig. 2. Wiring details for a programme card

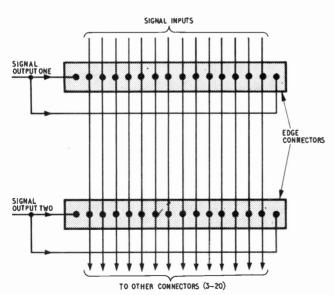


Fig. 1. Suggested wiring for the edge connectors

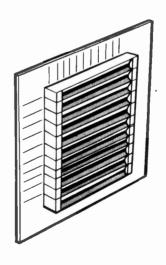


Fig. 3. Typical arrangement of connectors mounted on panel of synthesiser



ELECTRONICS — A handbook for Engineers and Scientists

By G. H. Olson Published by Butterworths 482 pages. Price £7

For increasingly the influence of electronics is felt by more and more people whose disciplines would traditionally have been unrelated. Probably the biggest contribution to this has come in the application of electronic techniques to measurement and control in various forms of sophisticated instrumentation. All of this has demanded an understanding of the subject, be it ever so superficial.

As a lecturer with an awareness of the "language" problem of the non-electronic specialist the author has provided a course which is qualitative and informative and will be prized as a valuable reference. The depth of subject treatment is not superficial nor does it lean towards being an examination textbook. In these, usually very little attention is paid to practical details in rigorous analyses. In the handbook these aspects are common and some useful circuits are given with details of the components.

From an introductory chapter on passive components, simple circuits containing them are analysed. Semiconductor devices are then examined (thermionic valves are eschewed throughout) followed by a chapter on indicating instruments. Next comes power supplies, then a massive chapter on the various genre of amplifier.

The concluding chapters embrace oscillators, an introduction to logic and digital circuits and measur-

ing instruments.

Useful references for further reading are included after chapters and appendices provide the means to resolve some of the finer points mathematically.

At £7 pricey, but a good investment for the d.i.y. tyro.

G.G.

RECORDING WITH COMPACT CASSETTES Published by Agfa-Gavaert 91-pages, 210mm × 150mm. Price 65p

THE THEORY and practice of cassette tape recording are combined in this concise, non-technical,

well-illustrated book.

Besides the cassette itself there is an excellent description of the theory behind tape recording, from the conversion of sound into an electronic signal to the actual recording of the signal by the head.

The book is full of hints and suggestions for getting the most out of a cassette recorder.

Many of the illustrations are in full colour, and overall the book seems excellent value for money.

This book is just one of a series on tape recording and photography and can be obtained direct from Agfa-Gavaert Ltd., Great West Road, Brentford, Middlesex.

S.R.L.

QUESTIONS AND ANSWERS ON INTEGRATED CIRCUITS

By R. G. Hibberd Published by Newnes-Butterworths 96 pages, 6½ in × 4½ in. Price 75p

A S ANY regular reader of Practical Electronics will realise, the integrated circuit is now a well-established component in modern electronics. However, the development of these devices from the unusual to the commonplace has been so rapid that many people will no doubt be feeling a little left behind.

For anyone in this position, this little book provides an excellent introduction to the field. In the form of questions and answers it gives the reader information about how integrated circuits are made, their advantages and disadvantages over discrete components, and their uses, both in the area of digital and linear electronics.

In the section on Digital Integrated Circuits there is a useful introduction to logic which is essential to the understanding of digital systems.

The book is written for students and technicians who require a simple, mainly non-mathematical, text on this rapidly widening facet of electronics.

S.R.L.

INSTRUMENT TECHNOLOGY (Volume 1) By E. B. Jones Published by Newnes-Butterworths Ltd. 402 pages, 9½ in × 6½ in. Price £5:00

THIS is a new (third) edition of a book which was first published in 1953 and forms the first volume of a three volume set dealing with Instrument Technology. The coverage of this volume is restricted to the measurement of pressure, level, flow and temperature and the treatment uses S.I. units throughout. The presentation is lucid and the mathematics have been kept as simple as possible. Together with the other two volumes of the set, this book should provide a good coverage of much of the material normally found in technician and craftsman level courses dealing with process control and instrument technology and should also prove useful as a reference text for engineers and other personnel who are not experienced in this field.

The text is well illustrated with clear diagrams, photographs and exploded views of numerous instruments which should prove useful to students especially, since they do not normally have access to such a comprehensive range of instrumentation.

Electronic devices and techniques such as the pressure sensitive transistor and the application of lasers to level measurement are not covered and the brief section on transistors, diodes and photocells does not reflect present day technology. In a work such as this the selection of material is difficult due to the wide range of instruments in current use. The author's stated aim has been to give a reasonably complete picture of the more common and more important devices in present use and this has been achieved. It is unfortunate that the price of this book will put it out of the range of the average student's pocket.

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4.10 miror amp meter. Large oblong, full vision front, size 4\frac{1}{2} in \text{ 3}; in approx. Good quality Japanese made, moving coil meter. \$2.65.

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RANNAGHT DISPLAY

By K.L. SPENCE

This article describes a Random Light Display which, unlike most light displays, is not driver by an audio signal.

The unit consists of nine or more coloured lamps positioned behind a semi-transparent screen. Each lamp flashes on and off in an apparently random manner, without an overall fixed sequence. The resulting effect is to fill the screen with a blaze of mixing and changing colours forming a fascinating display,

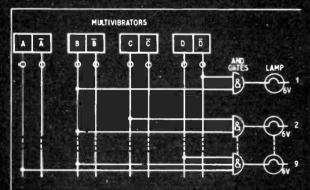


Fig. 1. Elock diagram showing the arrangement of the multivibrators and AND gates

METHOD OF OPERATION

The apparent random effect is achieved by using four slow-running multivibrators each with different frequencies and mark-space ratios.

Each lamp is driven by gating together the outputs of two or more multivibrators using a logic AND gate. The system is illustrated in Fig. 1, which shows the overall arrangement of the multivibrators and AND gates.

It can be seen that lamp LP1 will only light when multivibrator outputs B and \overline{D} are high, i.e. logic 1: lamp LP2 requires outputs B and C high, and lamp LP9 requires A, B and D high.

CIRCUIT DETAILS

The circuit diagram of a multivibrator is shown in Fig. 2. Four of these are required, the values of the timing capacitors C1 and C2 being given in the table.

The AND gate and lamp driver circuit is shown in Fig. 3

The gate may have two, three or four diode inputs. If any of the inputs is taken to 0V by a multi-vibrator output then current will flow down resistor

COMPONENTS . . .

MULTIVIBRATORS-4 OFF REQUIRED

Resistors

R1, R4 $5.6k\Omega$ (2 off) R2, R3 $33k\Omega$ (2 off) W 10% carbon

Capacitors

C1, C2 see table

Transistors

TR1, TR2 2N708 (2 off)

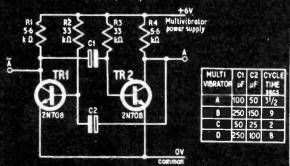
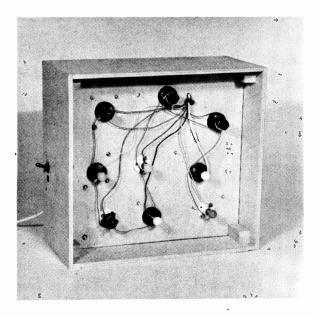


Fig. 2. Circuit diagram of a single multivibrator circuit. The table shows the values of the timing capacitors for the four multivibrators required



Photograph showing a front view of the completed unit with the cover removed

R1 and the voltage at TR1 base will not be sufficient to switch on transistors TR1 and TR2 which need about 1.4V to be forward biased.

However, if all inputs are at 6V then the diodes will be reverse biased and the current through R1 will flow into the base of TR1 causing it and TR2 to conduct and the lamp to light.

COMPONENTS . . . AND GATE/LAMP DRIVERS-9 OFF REQUIRED R1 8.2k Ω \pm W 10% carbon Diodes D1-D4 OA200 (4 off, or as required) Transistors TR1 2N708 TR2 BFY52 Lamps LP1 6V 0.3A m.e.s. with lampholder LPI 6V, 0-3 Å kΩ 0A200 0٧

Fig. 3. Circuit diagram of a single AND gate/ lamp driver circuit. Nine of these are required. The number of diode inputs may be two, three or four

COMMON

COMPONENTS . . .

POWER SUPPLY

Resistors

R1 82Ω ¼W 10% carbon

Capacitors

C1 500μF 12V elect.

C2 250 µF 12V elect.

Diodes

D1-D4 50V 5A bridge rectifier

D5-D8 50V \(\frac{1}{2}\)A bridge rectifier D9 6.2V 400mW Zener

Transformer

T1 Mains primary, 6-3V 3-6A secondary

T2 Mains primary, 6.3V 1A secondary

Miscellaneous

S1 2 pole mains on/off

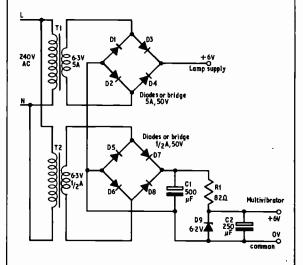


Fig. 4. Circuit diagram of the power supply. This provides a smoothed supply for the multivibrators and unsmoothed supply for the lamps

POWER SUPPLY

The power supply is shown in Fig. 4. It has two 6V outputs, one is smoothed for the multivibrators and the other is unsmoothed for the lamps.

The components specified are widely available, though discrete diodes instead of a bridge may be used if desired.

CONSTRUCTION

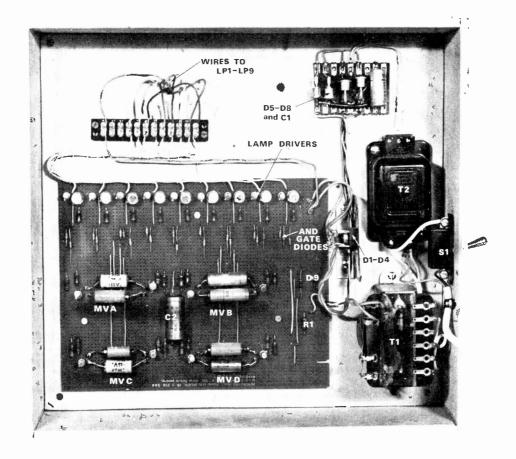
Component layout is not at all critical and the circuits may be built on any type of wiring board.

The author used a large piece of 0.1 in Veroboard, the general arrangement being seen in the photograph. Fig. 5 shows the multivibrator circuits and the lamp driver circuits in more detail. Four of the multivibrators and nine lamp drivers are required.

It is useful to provide a wiring pin for every multivibrator output and gate input as this greatly simplifies the cross-connections.

It is advisable to first construct and test all the multivibrators and gates and then make the connections between them.

The constructor is free to choose for himself which outputs to connect to each gate, the only restriction being that both the outputs of a particular



Photograph showing the general arrangement of the components at the rear of the box. The power supply Zener diode, resistor R1 and capacitor C2 are mounted on the Veroboard which carries four multivibrators (MVA to MVD) and nine AND gate/lamp drivers

multivibrator must not be connected to the same gate, as this would mean the lamp would never light.

Fig. 5. Suggested Veroboard layout for the multivibrators and for an AND gate/lamp driver circuit. The circuits shown are typical; breaks should be repeated as shown between circuits. The upper 6V line comes from the lamp supply and the lower from the multivibrator supply AMP DRIVER B € MULTIVIBRATOR

but it should be remembered that gates with four inputs will only be enabled very infrequently compared with the two-input gates. Photographs illustrate the general manner of con-

The gates may have two, three or four inputs

struction with lamps in the front of the box and the circuits behind.

The overall size is governed by the number of lamps the constructor decides to use. The prototype measured 15in \times 15in \times 6in and had nine lamps.

FRONT SCREEN

The front screen consists of a wooden frame covered with thin white cloth or tracing paper. A sheet of Perspex is then fixed on the front for protection.

The lamps may be any 6V types and they may be covered with coloured Cellophane or painted.

MODIFICATIONS

The unit described in this article may be extended to many more lamps and multivibrators providing the transformer can handle the power output.

Higher wattage lamps could be used by incorporating thyristors in the lamp driver circuits. It could also be interesting to have a mixture of AND and OR gates instead of only AND gates.

The values of the timing capacitors can be changed to give more rapid or slower changes as the constructor desires.

PE RORDO

PART f 8

E.M. STEREO TUNER

BY R.A. COLE

The Rondo system is designed specifically to accept a variety of inputs to suit individual constructor's tastes. However, the initial design concept included the basic idea of providing as many facilities as possible within the one unit. One of the most important to many constructors will of course be the stereo tuner and associated decoder and Part 8 will initiate the discussion of these two items.

GENERAL CIRCUIT

The basic system for an f.m. tuner and decoder is shown in block form in the accompanying Fig. 8.1 and the complete circuit in Fig. 8.2. A varicap-tuned head feeds its output to a wide-band amplifier, a ceramic filter, a further i.f. amplifier and finally to the stereo decoder. Movement of the varicap potentiometer VR1 is indicated on a suitable tuning dial and a tuning meter indicates signal level.

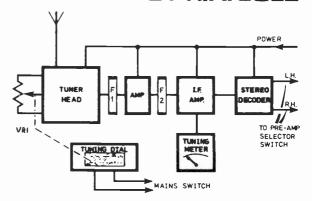


Fig. 8.1. Block diagram of the f.m. stereo tuner/decoder arrangement for the Rondo system

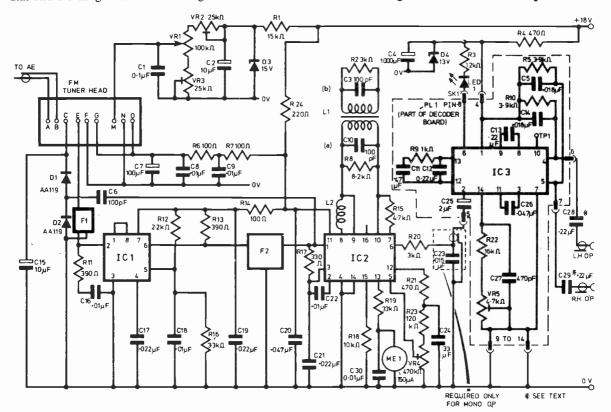


Fig. 8.2. Full circuit diagram of the tuner and decoder with the decoder board shown shaded

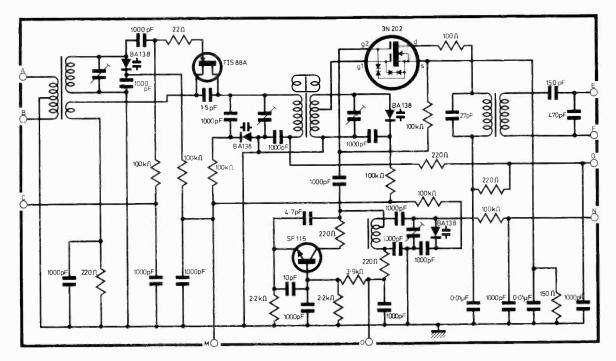


Fig. 8.3. Detail circuit of the Larsholt tuner head as used in the Rondo system

The tuning head used in the Rondo is a Larsholt unit, pre-built in Denmark, which makes use of j.u.g.f.e.t., m.o.s.f.e.t. and bipolar devices to give a high standard of performance

The head feeds via a ceramic filter F1, amplifier IC1 and filter F2 into the i.f. section. This latter is the monolithic amplifier TDA1200 or the equivalent CA3089E.

I.F. AMPLIFIER

In the i.f. amplifier there are three stages of amplification and associated level detectors that feed both delayed a.g.c. for the r.f. amplifier in the tuner, and the tuning meter. The amplification stages then feed the detector section.

The detector, a quadrature detector, requires tuning using one or two external coils and in practice this involves only peaking for maximum audio output. In fact, the use of one or two inductors is optional. A single tuned inductor L1 (a) gives distortion levels around 0.5 per cent whilst the use of a double inductor configuration L1 (a + b) allows levels as low as 0.1 per cent to be reached.

Audio output from the detector is fed to a further stage of amplification whilst, at the same time, a signal is fed to a level detector which in turn controls the mute drive circuits. This latter is connected externally of the chip to a muting sensitivity control and fed back to an audio mute control amplifier which is coupled to the audio output amplifier.

The quadrature detector also feeds an a.f.c. drive amplifier.

Finally, the output of the tuner is fed to a stereo decoder which in the present case uses the well-known Motorola MC1310P or the RCA CA1310E phase-locked loop monolithic integrated circuit.

The circuit of the whole system is shown in Fig. 8.2, the i.c.s and tuner head being shown in outline only.

TUNER HEAD

The tuner head circuit is shown for interest in Fig. 8.3. This Larsholt product was chosen for its sensitivity and the quality of output. Heads of this type are of course pre-aligned by the manufacturer and require no further attention from the user apart from perhaps some peaking of the output by tuning to suit the external circuitry. In the present instance this is limited to peaking of the output inductor connected to terminals E and F of the head.

The head and most of the associated circuitry are mounted on one circuit board which is located within the trough assembly of the Rondo directly beneath the pre-amplifier board but orientated parallel to the bottom of the trough.

The stereo decoder i.c., IC3 in Fig. 8.2, and associated components are mounted on a further board which is suspended under the fascia and master tone board to the top right of the assembly when viewed from the front.

STEREO DECODER

For ease of description we will consider one of the two alternative decoders as they are virtually identical.

The MC1310P is a phase-locked loop decoder with typical channel separation figures of 40dB and total harmonic distortion typically 0-3 per cent. It is carried in a 14-pin d.i.l. package.

This decoder does not require any inductors and in fact the only adjustment it might require in practice is the setting of the voltage controlled oscillator (v.c.o.). This is achieved by adjustment of VR5.

The chip automatically detects the presence or absence of stereo by detecting the presence of the 19kHz pilot tone at the input. The input circuit includes a threshold circuit which sets the detection of the pilot tone at about 16mV so as to stop the

STEREO DECODER

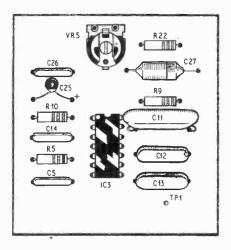


Fig. 8.4. Component layout on p.c.b. for the stereo decoder



Fig. 8.5. Master p.c.b. copper layout for the decoder board

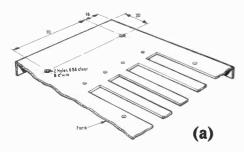
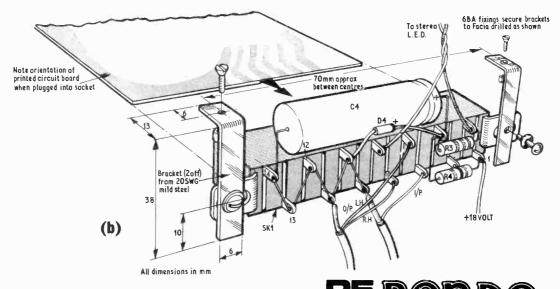


Fig. 8.6. (a) Rear right-hand corner of fascia panel (see Fig. 4.1). (b) Details for the mounting of the decoder board edge connector and associated carrying brackets, their location in the main assembly, and the positioning of some off-board components



COMPONENTS . . .

Re	sistors	3		
F	₹1	15k Ω	R13	390Ω
F	R2	$3k\Omega$	R14	100Ω
F	₹3	1·2kΩ	R15	4·7kΩ
F	₹4	470Ω	R16	3⋅3kΩ
F	₹5	3.9kΩ	R17	330Ω
F	₹6	100Ω	R18	10k Ω
F	₹7	100Ω	R19	33 k Ω
F	₹8	8-2kΩ	R20	3kΩ
F	₹9	1kΩ	R21	470Ω
F	₹10	3.9kΩ	R22	16kΩ, ₩, 2%
F	₹11	390Ω	R23	120kΩ
F	₹12	2-2kΩ	R24	220Ω
1	ΔII ∦W,	5% except where	stated	

Cá	n	а	•	iŧ	n	rs

Capacit	013		
Č1	0·1μF	C18	0·01μF
C2	10μF, 25V	C19	0.022µF
C3	100pF Polystyrene	C20	0·047µF
C4	1,000μF	C21	0·022μF
C5	0·018μF	C22	0.01µF
C6	100pF Polystyrene	C23	0·015μF,
C7	100μF, 16V		De-emphasis
C8	0·01µF	C24	0-33μF
C9	0·01μF	C25	2μF or greater.
C10	100pF	Tantalum	bead preferred
C11	0·47μF	C26	0.047µF
C12	0·22μF	C27	470pF,
C13	0·22μF		2% Polystyrene
C14	0·018μF	C28	0.22µF
C15	10μ F , 25V	C29	0·22μF
C16	0·01μF	C30	0·01μF
C17	0.022µF		•

Potentiometers

VR1	100kΩ Lin
VR2	25kΩ Lin preset
VR3	25kΩ Lin preset
VR4	470kΩ Lin. preset
VR5	4.7kΩ Lin. preset

Diodes

D1	A A 119	
D2	A A 119	
Da	D7V00	15\/

BZY88, 15V Zener, 400mW BZY88, 13V Zener, 400mW LED1 Light emitting diode

Integrated Circuits

IC1	CA 3053
IC2	CA 3089E or TDA 1200
IC3	MC1310P or CA1310E

Inductors

L1	Single quadrature, KACS K586 HM or
	Double quadrature TKACS 34343 AVO
	and TKACS 34342 AVO from Toko U.K.
L2	R.F. Choke, 22µH

Head Tuning meter MEI	Larsholt 8317-501 stereo
Tuning meter MEI	tuner head, f.m. 0-150µA Full-View panel meter
Ceramic Filters F1, F2	CFS 107M (Toko)

Assorted wire, co-axial cable, p.c.b., 14-way edge connector, metalwork brackets, nuts and screws

decoder switching to stereo mode when the audio is too weak to give good signal-to-noise ratio.

Indication of stereo presence is given by the light emitting diode LED1 which is wired in series with a $1.2k\Omega$ resistor. This latter drops the voltage across the diode to around 1 6V at 20mA. As most diodes give adequate light emission at this current this seems sensible even though the i.c. is capable of driving up to 100mA if necessary.

The decoder and associated de-emphasis components are mounted on one board as shown in Fig. 8.4. The master negative for this board is shown in Fig. 8.5 and two combined scrap views show the orientation of the board in its edge connector and the mounting of this latter under the master tone

control board in Fig. 8.6. Assembly of the decoder board follows normal p.c.b. assembly procedure although it will be noted that several components from the decoder and tuner-associated circuitry are in fact mounted in the wiring, as it were. Thus R3 and R4, C4 and D4 are all mounted on the edge connector carrying the decoder board. The l.e.d. is mounted in the tuner fascia just above the tuning meter. C28 and C29 can be mounted on the pre-amplifier board.

Wiring from the decoder board is passed down the side of the fascia to the fanning-strip/switch assembly described in the last part for interconnection to the remainder of the looming.

OSCILLATOR ALIGNMENT

When the tuner/decoder system has been completed it is necessary to set the decoder oscillator. This is a simple procedure requiring that one tunes to a known broadcast and then adjusts VR5 until the pilot light illuminates. This of course assumes a stereo broadcast is taking place.

Stereo output should now be heard from the

loudspeakers.

If an oscilloscope is to hand the adjustment can be made perhaps more accurately by connecting the 'scope to the v.c.o. test point TP1 (pin 10 of the i.c.) and adjusting VR5 until a 19kHz waveform (55 µs cycle time) appears. Equally, a counter/timer could be used to register 19kHz.

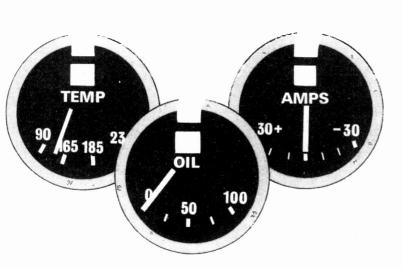
Tolerance of this value is ± 3 per cent or 580Hz either side of 19kHz.

I.C. HOLDERS

An important point is the use of i.c. holders with these types of i.c.s. It is always advisable to use holders since this gives the obvious advantage of being able to remove the i.c. later and in any case some manufacturers revoke their guarantee if their particular products have been soldered in position as there is always the risk of thermal damage occurring.

It will be appreciated that each additional item of equipment inserted in the Rondo is an extra load applied to the power supply. In general terms this is no problem since the p.s.u. is well able to deal with the demand placed on it. However, the problem is easily overcome if it does arise by simply turning TR6 of Fig. 3.1 in Part 3 "on" a little more. This can be achieved by simply lowering the value of R32 in the same figure to 470Ω .

Next month: Construction details for the tuner board and mechanical assembly



CAR SYSTEMS MONITOR

BY J.P. PERRY

EVEN in this day and age it is not unusual to find cars stranded at the side of a motorway after having suffered boiling perhaps through loss of a fan belt or a burst hose.

Remembering that most vehicles are fitted with some form of temperature indicator for water temperature this sort of event only serves to indicate just how little attention is paid to the various dials by a driver, particularly when coping with things like motorways.

Oil, water and indeed battery charge, are all indicated visually and their loss, which can occur quietly and seemingly unnoticed, can have very damaging results including seized engine bearings or the like. So clearly what is needed is some form of audible indication in the event of failure of these parameters.

The present circuit was designed to give an instantaneous audible alarm if water temperature, oil pressure or generator output assume a fault condition.

The circuit is fairly simple and would probably cost in the region of £3 which is cheap in comparison with the cost of replacing a damaged engine or even carrying out a minor repair with labour costs in the £3 per hour region.

PRACTICAL CIRCUIT

A circuit is shown in Fig. 1. Three inputs can be connected to detect the presence or absence of voltage at for example the car temperature gauge, the ignition warning light and the oil pressure switch. In each case if the respective point is connected to chassis (earth) an alarm condition can be said to exist.

If this occurs current is taken by the associated one of the three diodes D1, D2 or D4. TR3 becomes saturated and 12V is applied to the astable TR4/TR5. Protection diodes D6 and D7 prevent the breakdown of the base/emitter junctions of these latter devices.

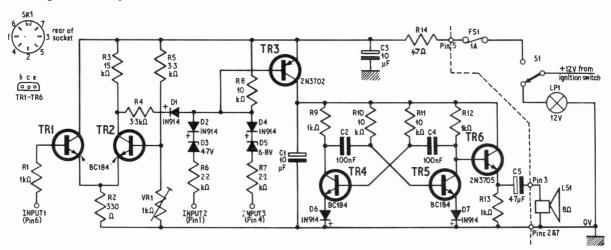


Fig. 1. Circuit diagram of Monitor

The astable is connected to a final transistor TR6 to drive a loudspeaker which may be any 2 to 3 inch 3 to 30 ohm device. The level of output is variable by choice of a suitable value for R13, $1k\Omega$ being a suitable value to start with.

LAYOUT

Veroboard is used for the layout and cutting pattern and component locations are shown in Fig. 2.

No particular problems exist as far as the board is concerned except perhaps with the diodes with which care should be taken when bending the leads to avoid cracking the glass bead.

Whilst the circuit is shown for negative earth vehicles, the positive earth version can be catered for quite simply. It only requires that the *npn* transistors be replaced with *pnp* and the *pnp* with *npn*, that the diodes be reversed and the electrolytic capacitors similarly turned round.

Whilst BC184's are specified together with 2N3702's and 1N914's almost any small signal transistors and diodes can be used.

TESTING

Before wiring in to a vehicle it is best to check that all is functioning properly. Simply apply the 12 to 15V supply and temporarily connect input 1, the water temperature gauge input, to +12V. There should be no output.

If now any one of the inputs are connected to ground, an output should be obtained at the loud-speaker.

The volume of this buzz can be varied by selecting R13 to suit requirements.

CASE

Thus far we have only discussed a Veroboard assembly which, for use in a vehicle, should obviously

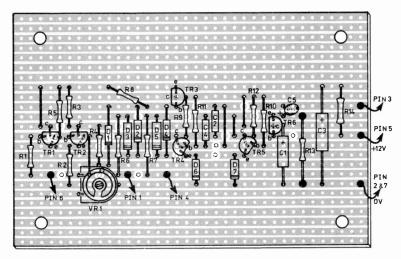
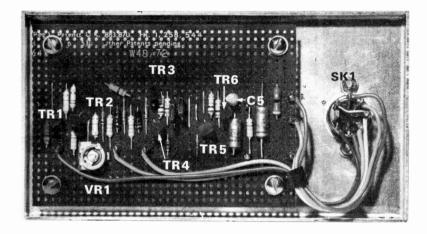


Fig. 2. Component layout on Veroboard, the copper track cuts being shown as white cut-outs

COMPONENTS . . .

Resis	tors			Transistors	
R1	1kΩ	R8	10k Ω	TR1, TR2, TR4, TR5	
R2	330 Ω	R9	1kΩ		BC107, BC108, BC109
R3	15k Ω	R10	10k Ω	TR3	2N3702 or 2N3703
R4	3-3k Ω	R11	10k Ω	TR6	2N3705 or 2N3704, BC182
R5		R12	1kΩ		BC183, BC184, "L" types
R6	2·2k Ω	R13	1k Ω (see text)	Note that when select	cting alternatives care should
R7	2.2k Ω	R14	47 Ω	be taken with the pin co	onnections.
All a	W carbon film	1			
				Diodes	
Poter	tiometers			D1, D2, D4, D6, D7	1N914 or 1S914, 1S44, 1N4148
	1k Ω preset			*D3	BZY88 4.7V Zener, 400mW
	Tital product			D5	BZY88 6-8V Zener, 400mW
Capa	citors				
	10µF electroly	tic 16V		Miscellaneous	
C2 100nF ceramic				SK1 6-pin DIN.	
C3 10μF electrolytic 16V C4 100nF ceramic				Veroboard, Veropins	s, aluminium box, connecting
				wire, fuseholders,	switch and lamp if required
C5	4.7µF Tant. be			solder, etc.	



Interior view of Systems Monitor showing completed unit

be cased in some form of protective box. In the prototype this took the form of a simple aluminium case and lid, with the Veroboard mounted on support pillars in the lid.

Wiring access to and from the circuit is provided via a 6-pin DIN socket SK1 mounted in the lid. As the whole assembly of lid and board is only a matter of an inch deep clearly the box can be quite shallow but it is best to use a box which can be screwed up and properly sealed against ingress of moisture from the engine compartment or wherever it is located on the vehicle.

INSTALLATION

It is probably wisest when installing any equipment in a vehicle to first of all disconnect the battery so as to avoid any damage from inadvertent shorts. Power supply to the unit is connected from the ignition switch, switched side and the inputs 1. 2 and 3 are connected respectively to the water temperature warning light or thermometer, the oil pressure warning light or block switch and the

generator output terminal or terminal "A" on the ignition light.

The battery may now be reconnected and no output should be obtained.

If the constructor has wired in the extra components S1, FS1 and LP1, then when the ignition is switched on LP1 should come on but there should still be no output. Switching S1 to the "on" position should extinguish LPI and the unit is now "armed" for operation. In fact an ouput should be obtained because of course there is no oil pressure or generator output.

Starting the car should silence the output but if this fails it may be necessary to adjust VR1.

The vehicle should now be driven round for a while till engine temperature reaches its normal level and VR1 is again adjusted till the output buzz is just extinguished. This is all the setting up required with the possible exception of small adjustment to make up for the difference between summer and winter temperatures.



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NEWS BRIEFS

OBITUARY

HE name Bulgin must be one of the best-known in the electronics field in Britain. Certainly every follower must have used plugs, sockets and so on bearing this name.

Thus it is with great sadness that we mourn the passing of Arthur F. Bulgin, O.B.E. the founder of the company bearing his name. He has served our industry for 51 years and, indeed, his country as well, and it is a mark of his astute judgement that the A. F. Bulgin company is the only one in the industry to show a steady rate of growth since the last war.

Chairmanship of the company continues in Mr Bulgin's son, Mr Ronald Bulgin, M.A., C.Eng., M.I.E.E., who also shares joint managing directorship with A.F.B.'s brother, Mr Stanley Bulgin.

....tuning in to ESP

Our correspondent, Peter Verwig, recently attended the IEEE Intercon show in New York. Among the usual mass of technical and marketing papers read at the conference was a session on "New Advances in Parapsychology".

The chairman of the technical programme committee, Jack Raper, was one of the doubters on the advisability of including a parapsychology session in the International Convention and Exposition of the Institute of Electrical and Electronic Engineers in New York. In the end, the session went ahead and, if not exactly a sensation, it certainly caused a stir among the coolly-calculating, logically-minded scientists and engineers who formed the audience.

LECTURERS

The one aspect in favour of the session was that though the speakers may have had odd ideas, some of which might be classified as kinky, they did at least have the credibility of being engineers with normal levels of engineering objectivity and were clearly not the crankier type of mystic living on emotion rather than scientific fact.

Two of the speakers, E. Douglas Dean and John Mihalasky were from the Newark College of Engineering and the third, James B. Beal, hails from NASA, Huntsville, though he was careful to point out that he was speaking as a private person and NASA was in no way involved.

All three speakers were firmly convinced that such phenomena as extrasensory perception (ESP), precognition and telepathy were not only possible but also capable of detection and measurement in the scientific sense although much remains to be done.

TELEPATHY

The main practical work of E. Douglas Dean has been in telepathy. For his experiments he uses the plethysmograph, an instrument used in medicine for over a hundred years. It records a change in volume with correlation against blood circulation in the extremities of the body, for example the finger tip.

In the ordinary way a person has no control over any variations, but it has been shown that what we might loosely term "emotion" causes involuntary vaso-constriction leading to a lowering of blood volume in the measured part of the body. Dean's apparatus is considerably more sensitive than the old medical

plethysmograph and incorporates a pressure capacitance diaphragm pick-up and a continuous chart recorder which records the normal pulse of the heartbeat and any vaso-constriction as a pronounced dip below the average baseline.

The "subject" lies on a couch with a finger connected to the instrument and in an adjacent room the "agent" picks up a card on which is inscribed a name which may or may not have significance for the subject. Some cards are blank. The subject doesn't know what is on the cards, or the order in which they are shuffled.

The agent looks at each card for twenty seconds. Blank cards showed no response and cards with names only showed subject response on the recording pen when they had some emotional significance, for example a close relative, and the pen deflection was most pronounced when there had been recent anxiety or other strong emotion in connection with the name.

In one test the subject's two-day old new-born baby's name was on the card and the resulting vaso-constriction was so severe it took over a minute for the recorder to return to baseline.

The experiments were repeated with the agent enclosed in a fully electrically screened room and this had no effect, indicating that whatever the communication medium between agent and subject it was not affected by screening. Long distance tests have been completed north to south from New York to Florida, a distance of 1,200 miles, and eastto-west from Newark, New Jersey, to Bordeaux, France. Amazingly, the effect was also recorded with the agent 35ft under water in Florida and the subject 4,000 miles away in Zurich, Switzerland.

The underwater test opens up the possibility of communicating with submarines submerged at great depth, of greatest potential value in the event of disabling disaster. The technique would involve the use of pre-arranged time slots and the Morse code with stimulus, for example, representing a dot and non-stimulus being a dash. The data rate would be very slow but over a period of hours it is thought that comprehensible communication could be possible.

An important finding by Dean is that using electronic equipment to establish emotional response shows scientifically that one person in every four has some level of ESP ability.

FIELD EFFECTS

James B. Beal maintains that electric fields have important effects on the body. He has gone as far as to fit a 500V positive field generator in his car to alleviate fatigue. He has also reared beans in a 2,000V positive electrostatic field and recorded a four-day earlier germination compared with a control group outside the field.

The same apparatus was used on a badly asthmatic child with beneficial results. Tests on typists showed an enormously lowered error-rate if they sat in a strong field.

Beal recalls that the ancient Chinese knew, perhaps instinctively, that beneficial health, mental and spiritual qualities could be obtained by proper geophysical orientation of homes and religious shrines in relation to the earth's magnetic poles, subterranean streams, geology and topology. It is even suggested that many mental patients are not ill but merely hypersensitive to field effects and, of course, it is well established that natural phenomena such as hot winds containing an excess of positive ions, especially in the Middle East, induce both physical and mental maladies in the population.

It was also suggested by Beal that ESP can be developed in suitable subjects by teaching machines. One such is commercially available.

SEEING THE FUTURE

Working on precognition (seeing into the future), especially in the realm of creative ideas in engineering and business has been studied for many years by Dr. John Mihalasky.

The successful engineer or businessman is often said to have intuition or insight in addition to ordinary levels of logical thought. The brilliant solution to a problem is often conceived "in a flash" from the unconscious mind spontaneously and frequently at very odd times and in peculiar circumstances.

SOLOP POVO and its applications for TODOU

The oldest known power source is the sun and there have been many ways devised to use its energy. Until very recently none have withstood the march of technological progress. Modern technology is now able to provide ways of making use of a very small fraction of that power that is always available.

The utilisation of solar energy has also been limited by economic reasons; the techniques for many applications have been known and fully proven but the cost of using them has made it totally uneconomic.

PHOTOCELLS AND SATELLITES

The basic component for the conversion of solar power into electrical power is the photocell and this was introduced, as a production device, in 1957.

This cell, in the form of a silicon photovoltaic semiconductor diode, has been the prime power source on earth orbiting satellites since the first was launched in 1957. It is therefore not surprising that industry is now using this well established, fully proven space technology as a source of pollutionless terrestrial power.

GROWING SILICON CRYSTALS

The photovoltaic cell is produced from silicon, the second most abundant element on earth.

A seed of pure single crystal silicon is dipped into a bath of molten silicon and rotated and slowly withdrawn, in an inert atmosphere, producing a bar of almost pure single crystal silicon.

The silicon is required to be doped with small quantities of impurities to control its conductivity. These impurities are introduced into the bath of molten silicon so that the withdrawn bar contains the correct level of dopants.

Early technology imposed a limit on the diameter of bar that could be pulled of about one inch, but recent developments have seen this increased very significantly, and it is now usual for 2\frac{1}{2} in diameter bars to be used.

The bar of silicon is then cut so that circular slices of defined thickness are produced. The majority of cells produced for satellite applications have used silicon slices 300 micron (one micron = 10^{-6} metre) thick but a new generation requires cells of 200 micron thick. The thickness of silicon used does not affect production processing.

Almost all silicon produced for cell applications is boron doped so that it will conduct by means of positive charges or holes, and for this reason is called p-type.

PHOTOCELL PRODUCTION

The production of a photocell involves several basic steps and these may be listed as follows:

- 1. Diffusion of pn junction
- 2. Contact formation
- 3. Coating

The creation of the pn junction is achieved by the controlled diffusion of an impurity into the p-type silicon slice. The impurity used is phosphorus which is introduced into the diffusion furnace as a vapour, the depth of the diffusion being a function of time and temperature.

The basis of operation of the photocell is such that when photons (light particles) strike the cell surface, close to the pn junction, positive charges (holes) and negative charges (electrons) are produced. The electrons will tend to travel towards the n-type silicon and the holes towards the p-type silicon. If connections are made to the n- and p-regions electric current can flow into an external load.

METALLISED CONTACTS

The next vital production stage is thus the deposition of a metallised contact structure. The p-contact, on the rear face of the cell, can cover the total cell area, but the design of the n-contact, on the front or active surface, is critical to the performance of the cell.

^{*}Electronic Components Division, Ferranti Ltd.

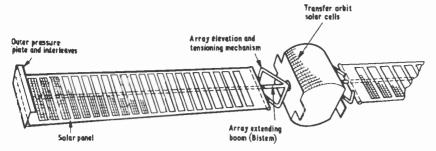


Fig. 2. Drawing of a flexible fold-up solar array. Each paddle is about one metre wide by five metres long

The requirements are many and varied and, in some cases, conflicting. For maximum cell output the contact must mask minimum cell area but must have sufficient cross-sectional area to handle the photon generated current.

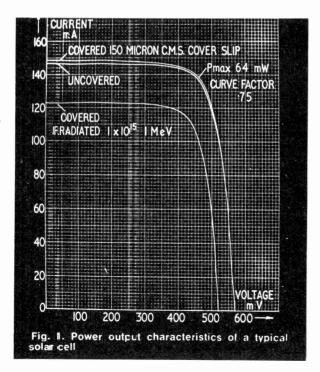
Thick metallisation must be avoided as severe stresses can be set up on thermal cycling and, in the extreme, this stress could be sufficient to break the silicon. The metallisation must have good electrical and thermal conductivity, must be easily contactable, and should not degrade or cause degradation in any encountered environment.

The process by which the metallisation pattern is defined must also be of sufficient accuracy that overall pattern tolerances are maintained. Pattern definition is achieved by a photographic mask and the metallisation used on the satellite cell is a composite structure of nickel, copper and gold.

The nickel gives the ohmic contact to the cell, the copper gives the electrical conductivity and the gold

prevents contact tarnishing.

The pattern defined on the satellite cell is a narrow top contact bar and then 24 narrow fingers running down the length of the cell. These fingers are typically 25 microns wide and five microns thick. The multiplicity of these fingers optimises the collection efficiency of the photon generated electric current.



OPTIMISING CELL PERFORMANCE

On bare silicon, light striking the active cell surface will be partially absorbed but a certain percentage will be lost due to reflection. This percentage can be significantly reduced if a quarter wavelength thick transparent dielectric layer is deposited on the cell surface.

The refractive index of this layer is chosen to optimise cell performance. Early cells used either silicon monoxide or dioxide but the latest technology uses titanium oxide.

The almost universally used cell size for satellite applications has been 2 cm sq and the output power from a cell of this type is typically 62mW at 25°C when illuminated with air mass zero sunlight at 140mW/cm². This gives an overall power conversion efficiency of 11 per cent.

Fig. 1 shows the output characteristics of a typical solar cell.

The power requirements for a satellite have always demanded a cell of maximum power, from minimum area, at minimum weight and this has dictated that the cell produced to satisfy this requirement is of the highest possible performance.

SATELLITE SOLAR ARRAYS

Satellite solar array design has followed two distinct channels, both of which offer advantages and disadvantages in the satisfying of an overall system requirement.

In one case the cells are mounted on a cylindrical body which spins with its axis perpendicular to the sun. In the other case the cells are mounted on deployed paddles which are sun orientated.

To give some idea of sizes involved, the recently launched communications satellites designated *Intelsat 4*, each carry 50,000 cells mounted on a cylindrical body, nine feet in diameter and ten feet high.

This initially will provide almost 900W of power and will provide power for 5,000 simultaneous telephone channels or ten colour television transmissions. To provide 1kW of power from a deployable array would require approximately 17,000 cells and, using two paddles, each paddle being typically one metre wide and five metres long (see Fig. 2).

THE ADVANTAGES OF SILICON

The justification for the choice of silicon is simple. As previously stated it is the second most abundant element and it has been the foundation of the semiconductor industry. As such it has seen an investment in its technology many orders greater than all other basic semiconductor materials.

The basis of using the experience gained from developing and producing solar cells for satellite

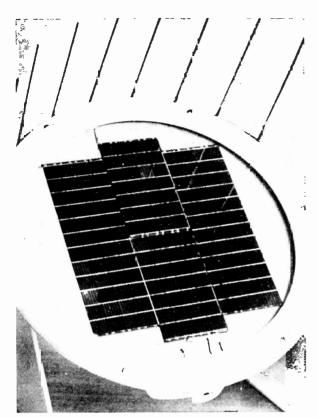


Fig. 3. A terrestrial solar panel as used for marine applications

applications is that the silicon voltaic cell has been the prime power service in over 700 American and some 500 Russian spacecraft of varying type.

From this background the selection of a silicon photovoltaic cell as a design start point for terrestrial power applications was inevitable and inarguable.

POSSIBLE ALTERNATIVES TO SILICON

It should not be thought that silicon is the only material suitable, and some effort has been devoted to other semiconductors.

The most notable of these have been gallium arsenide and cadmium sulphide. The former is able to withstand high temperatures and thus is able to generate at intensities 1,000 times greater than sunlight. The problems of heat dissipation must still be solved, however, before it can be considered to be technically competitive, but on cost it must be extremely doubtful if it can ever be considered to be an economic alternative.

Cadmium sulphide offers a cheap means of producing a photocell of considerable area in a flexible sheet form. It does, however, suffer the disadvantage of much lower conversion efficiency, and hence will only ever be able to be suitable for that market segment that can accept no restrictions on area.

TERRESTRIAL SOLAR POWER

The initial choice of cell for terrestrial use was dictated by availability, and 2×2 cm reject satellite cells were used. These were cells that failed to fully satisfy the most stringent electrical and physical requirements of a space satellite specification but

which would work quite happily when mounted on a terrestrial panel.

The panel was designed to provide 400mA at 15V in the equivalent of bright sunlight, and required for this output 156 cells interconnected as 39 cells in series with four cell strings in parallel.

MARINE ELECTRONICS

The design and construction of this panel, ideal for off-shore repeater or navigation aids, is not really suitable for the larger volume commercial business. A fast expanding market is that of marine power generation. The use of instrumentation and electronics on yachts and power boats is ever increasing with very little regard for the state of the onboard battery.

A small solar panel permanently mounted on the boat is extremely attractive from an economical and environmental standpoint. The use of solar panels in marine applications will, however, be limited by price considerations and only when a panel can be produced at a low enough price will full market penetration be achieved.

LARGER SLICES

The techniques for the production of refined single crystal silicon have not stayed stationary and manufacturers now have readily available large area slices. This realises a slice diameter of typically 6cm and allows the use of cells of this size and proportional power rating.

Based upon expected conversion efficiencies the output from a circular cell of diameter 6cm should be 500mA at 470mV when measured under 100mW/cm² simulated sunlight at 25°C.

The front active cell surface will have a simple metallised contact pattern and the back face will be fully metallised. The interconnection of cells must be performed simply, easily and reliably, to permit the building of the basic power modules.

Based upon the major market requirement for recharging nominal 12V batteries the voltage capability of the module is defined as greater than 13.8V. Thus a basic power module can be defined as 30 cells interconnected in series.

The expected output from the package under simulated bright sunlight conditions will be 500mA at 12V. The calculation for overall charging rate must be subject to some hypothesis as to expected weekly hours of sunlight and at what illumination intensity.

AVERAGE SUNLIGHT

The average daily hours of daylight during summer can be considered to be 15 hours. The Meteorological Office estimate for hours of sunshine on the South coast of England is a daily average of 7 hours. Assuming this to be at 75 per cent intensity, this is equivalent to 5.25 hours at full sun.

Thus the daily output charge from the panel will be 2-6amp-hours. Based upon a five-day week for charging and the boat being in use the other two days this gives an average weekly charging capability of 13amp-hours, neglecting any charging that can still take place during the time the boat is in use.

The expected current drain, based upon estimated usage of electronics and ancillary electrical equipment, is of this same order, 10 to 15amp-hours, so the projected panel size would seem suitable for this market.

OTHER APPLICATIONS

Usage of this type of panel is by no means limited to marine applications. A similar application, and one that is being studied closely, is for use with caravan trailers.

The power output requirements for a solar panel in this application are almost the same as on board a boat and hence the market can, hopefully, be catered for with an identically produced product.

Remote or unmanned automatic operation of road signs is receiving great attention, especially in the use of danger signs for such hazards as road works. The often seen flashing amber lights denoting road repairs could be easily solar powered.

Built-in rechargeable batteries could be charged during daylight hours for operation at night. The use of this type of power source could easily be extended to cover motorway hazard lights, emergency telephone systems and many monitoring devices.

Solar power gives to these applications the advantage of freedom from main power links and must improve their suitability for utilisation in many more wide ranging areas.

FORESTRY AND METEOROLOGY

These applications are ones that are commonly encountered by a wide cross section of the public but extensive markets exist in less well known areas. Forestry requires extensive use of remote automatic equipment for such applications as firewarning and portable or remote communications. The use of this type of power source is also available for weather monitoring, rain gauges, flood warning, seismic detection. The list can be almost endless.

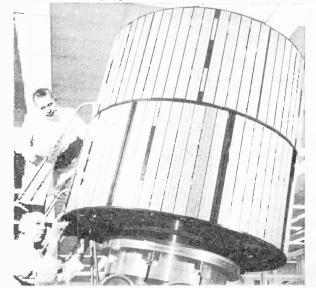
Two application areas that will demand a high reliability factor are those of remote area repeater links for communications and off-shore navigation

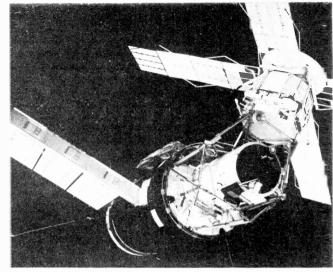
aids.

REPEATER STATIONS

Microwave relay stations and telephone repeater stations are often situated in some of the remotest parts of the world and are required to remain in unattended trouble-free operation for periods of a year or longer. It can cost a great deal of money to transpert people and spares to these sites for servicing and repair and it is therefore essential that frequency of visits is as low as possible.

Photograph of a satellite built by the Hughes Aircraft Company for NASA and launched in 1966





Photograph taken from Skylab 3 showing the deployed solar arrays

POWER PRODUCTION

The present major energy crisis has spurred even greater discussion and thought towards ways of producing, economically, consumer electrical power, such that it can be considered a realistic alternative to power produced from fossil fuels.

Two ideas, once considered close to science fiction, are being discussed. One involves a large solar farm in a flat desert region and the other a large solar array in synchronous orbit above the earth beaming r.f. power to a terrestial station.

Power would be taken away on a National Grid type system. The problems of maintaining the solar blankets can be imagined and no acceptable solution has been found.

ORBITING SOLAR ARRAYS

The orbital power station having two large arrays, totalling 7.3×2.7 miles, is the current design thought and these would flank a microwave transmitting antenna that would be 0.63 miles in diameter.

It is intended to try and fly a prototype satellite in 1990 and a commercial version could be flying at

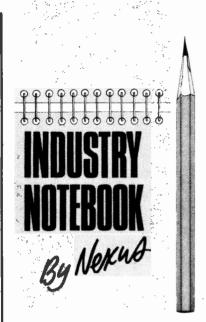
the end of the century.

THE SOLAR POWER AGE

The world markets for small commercial systems currently demand a high technology product in large volume at low cost but this is only the smallest tip of some vast technological iceberg. A technological miracle is in the making and it is one that could revolutionise our outlooks.

The photovoltaic cell and the solar array will become everyday commodities known to all. They offer a method of saving our precious and dwindling stocks of fossil fuels for processes that require these and they will provide an economic pollution free energy solution based upon a natural power source that will never be diminished.

We have seen the Stone Age, the Iron Age, the Industrial Revolution, the Space Age. The possibility of a Solar Power Age must not be dismissed or taken lightly.



WHAT'S IN A NAME?

We are all certain to be subjected during the next few months to growlings, grumblings, even open hostility directed towards the so-called multinational companies. In fact the very word multinational has now joined other political jargon words like racist, fascist, red, speculator, capitalist, terrorist in a fully emotive sense.

Multinational, in short, has become a word of abuse. Nothing to do with injection of capital and creating employment and wealth in underdeveloped or otherwise needy areas, and everything to do with faceless (indeed, probably heartless as well) exploiters counting their gold in plush offices thousands of miles away. No wonder some of the great multinational companies are keeping a low profile and, for one or two of them, with very good reasons.

Most of the outcry is directed against US-based multinationals but it has to be remembered that nearly all the great British companies operate multinationally, as do those of France, Germany and Italy.

I rather liked the argument put forward recently by the German Siemens group to an American audience in defence of their penetration of the US market by acquisition of US companies as well as direct sales from Germany. Siemens, said their spokesman Bernard Mayer, is a World Company and this has been true ever since the company was founded by Werner von Siemens, 126 years ago. He quoted the installation by Siemens of the 7,000 mile London to Calcutta telegraph completed in 1870 as an example of operating on a world scale. And he pointed

out that Siemens' advanced technology products, which demand two million dollars R and D expenditure alone every working day, could not be supported by domestic sales alone.

In the United States, Siemens has now grown from a handful of market research men 20 years ago to a sales and production force of 2,000. Steady, rather than dynamic growth, and still only a drop in the vast ocean of the US electronics and telecommunications industry.

My own rough guide lines for goodness factor in multinationals operating in Britain are that they should be British-managed, have good export potential, have some element of in-house R and D (i.e. not be iust an overseas production unit), and make a genuine contribution to the economic and social life of the country. Fortunately, there are more companies like this than of any other sort.

I suppose Mullard (Dutchowned) is the most notable example and of the more recently established companies, Hewlett-Packard (US-owned) has an impeccable record. But there are many others and it will do us all good when confronted with mindless political slogans to pause and think.

CUTTING THE COST

Electronics, because of its steady advance in solid-state technology, is uniquely running against the tide of rising world prices. You get more value per unit cost every year. The electronic watch which last year cost £100 will cost you £30 by 1978 according to industry forecasts.

It's all to do with volume. Last year solid-state watch sales were 250,000 in a world market of 185 million watches, only 0.14 per cent. But by 1978, say the pundits, in a projected market of 227 million watches, the solid state watch will have 51 million sales or 22:46 per cent of the total market. By that time the cost of the C/MOS LSI chip will have dropped from 10 US dollars to 2.50 dollars, the digital display from 9 US dollars to 2 dollars, the quartz crystal from 3 US dollars to 2 dollars and even labour (presumably through automation) from 5 US dollars to 2 dollars.

Take a single-chip electronic calculator. In 1970 the semi-conductor manufacturers made an "introductory offer" at 30 US dollars per chip. Last year, the price of the chip had plummeted to 5 dollars. This year the price is 4-50 dollars and the l.e.d. display is now 50 cents a digit instead of a dollar. The complete handheld calculator, albeit of the simplest type, is expected to fall

in cost to 20 US dollars by 1975, well under £10 at UK prices, the limiting factor being promotional costs because intense competition encourages heavy expenditure on advertising.

The cost of a single planar transistor in 1959 was 50 US dollars. By 1976, when it should be possible to buy the cheapest calculator chip of some 3,000 transistors for 3 US dollars, the cost per transistor will work out at 0.001 of a US dollar. That's not the end of the story because if all these transistors were discrete and had to be hand-wired with external wiring at a cost, say, of 10 US cents a wire installed, then the total cost would be astronomical.

The big question now is whether the semiconductor wizards can achieve the same order of cost reductions in solar cells for commercial and domestic use to alleviate the energy crisis? One line of research is to fabricate the cells in continuous ribbons instead of individually, but even the optimists concede that it may be another ten years before a hundredfold reduction in cost is achieved. This is the order of cost reduction needed to make economic the direct conversion from solar to electrical energy.

INSPIRATION

We could all kick ourselves at times for not spotting the obvious. The trembler bell in a telephone has been around ever since the telephone was patented by Alexander Graham Bell in 1876. If ever a mechanism needed up-dating it's that "oldy" in the base of your phone.

Mitel Canada Ltd. have come across with the answer, and it's an answer with enormous market potential. A replacement panel incorporating a 2-inch loudspeaker, amplifier and control circuits.

The all-electronic bell gives a pleasant ringing tone, as good or better than any electro-mechanical bell, and it has the advantage that it can also be used for ordinary loud-speaking messages. On a PABX system in a hotel, for example, the operator can make an emergency announcement to all rooms simultaneously or individual announcements to selected guests without the receiver being off its rest. If you are using the phone at the time, then the announcement automatically breaks into your conversation.

It is claimed that the Mitel "Mitone" signalling system can be manufactured to sell at the same cost as the old-fashioned single-purpose bell. Another advantage is that drive voltage is much lower and more in line with the requirements of modern electronic switchboards.

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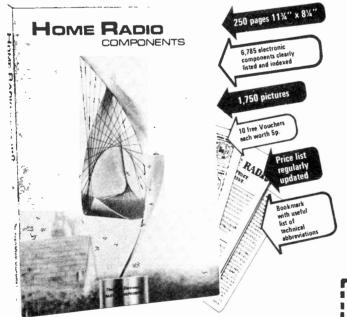
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MARCONI CENTENARY EXHIBITION

T O MARK the centenary of Guglielmo Marconi's birth, the Science Museum in Exhibition Road, London, is holding a special exhibition which will run throughout the summer to about October.

The exhibition consists of a large selection of documents, and a display of early apparatus together with many of Marconi's personalia and commemorative

material

Experiments with radio waves were carried out in 1888 by Hertz and were extended by various physicists in following years. It was not until 1894 when Hertz died. that the young Marconi, 20 years of age, appeared on the scene

After reading a commemorative article on Hertz he began experimenting in the attic of his parents' large villa near Bologna, Italy. By the summer of 1895 he had transmitted a signal over a few yards using a simple spark gap transmitter, and coherer type detector.

By August 1895 Marconi had achieved transmission over 14 miles. The Italian government remained unimpressed and so Marconi set sail for England.

Following preliminary demonstrations in his laboratory the chief engineer of the Post Office. William Preece, to whom Marconi had been introduced by his cousin, arranged formal demonstrations to Post Office officials.

Observers were suitably impressed and in a second demonstration six months later the range was extended to seven kilometres. An historic lecture at Toynbee Hall followed this and Marconi soon became a celebrity

The scientific community was, however, less enthusiastic suspecting Marconi of plagarism. In particular, Professor Oliver Lodge, who first developed the coherer and had in fact transmitted Morse signals over 60 metres before

Marconi began his experiments, was deeply offended.
In 1897 Marconi established "The Wireless Telegraph and Signal Company Limited" much to the disgust of the Post Office who did not favour experiments being carried out without their involvement.

In 1899 cross-channel transmissions were successfully achieved and the possibility of transatlantic radio was

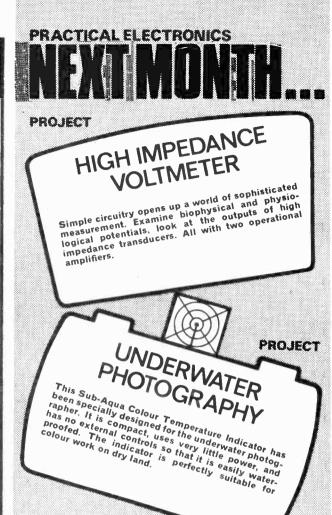
soon realised.

On the 12th December, 1901 the first transatlantic message, a single Morse letter S, was transmitted from England to Newfoundland giving Marconi success against all the odds.

Photograph of Marconi in London in 1896, showing the original apparatus brought by him from Italy. (Marconi Company Ltd.)



Practical Electronics July 1974



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IM is a game for two players. At each move a and only one, of several piles. He must remove at least one stick and the winner is the person to take the last stick. In an alternative form of the game it is the loser who is forced to take the last stick.

The game reputedly comes from ancient China, but was first analysed at the beginning of this century at Harvard, U.S.A. Clearly one or other of the players must win-there can be no draw-and it transpires that the winning strategy is closely related to the binary representation of the number of sticks in the piles.

Various designs for Nim playing machines have appeared over the years since the development of the digital computer and this article presents a contemporary one using readily available TTL digital integrated circuits.

WINNING STRATEGY

The prescription for winning at Nim relies on the fact that any arrangement of piles may be categorised as either a winning or a losing position. Further, if the arrangement represents a winning position the next move is certain to create a losing position, whilst it is always possible to play so that a losing combination is converted to a winning one

Thus if player A starts with a winning position, player B is bound to leave a losing one after his move. At this point A is able to convert the position to a winning one by making the correct move. The game is now back where we started and will continue

to cycle in this way until A finally wins.

Should A make a wrong move at any stage the way is open for B to take over the winning strategy. As there are many more losing combinations than winning ones it is unlikely that even a player who commences with a winning position in his favour should make all the moves appropriate to his being sure of winning.

BINARY REPRESENTATION

To discover whether a given arrangement represents a winning position or not we must first write down the numbers in the piles in the binary notation. For example, suppose that there are three piles containing 3, 6 and 7 sticks then we can write these numbers in binary form as three columns

Each column of this binary set should now be summed so that in this case we find the numbers 2, 3, 2, as shown. Now, a winning combination is one in which each power of two is represented an even number of times. Hence our example is a losing position because the second column, indicating the number of times that 21 occurs, adds up to 3 which is an odd number.

To convert such a losing position to a winning one a player must remove sticks from any pile containing the highest unpaired power of 2. In this example the highest unpaired power is 21, the middle column, and as each pile contains this power it is possible to produce a winning combination by taking sticks from any pile. In fact the combinations 1,6,7 3,4,7 and 3,6,5 are all winning positions.

Should the players choose the alternative game in which the last player loses, then the above strategy follows through unchanged until no pile contains more than one stick. In this case a winning position is one in which there is an odd number of piles of one match each. It will be seen that this is the opposite condition to that imposed by the normal game in this case.

BASIC DESIGN

Fig. 1 presents a block diagram for a machine which will implement the principles outlined. Rotary wafer switches represent the numbers of sticks in four piles 1 to 4, containing up to seven sticks each. These switches give outputs in binary coded form, so that, for example, if switch S1 is set to 5 then A_0 will be a logical 1 representing 2° , A_1 will be a logical 0 representing 2^1 and A_2 will be a logical 1 representing 2^2 . For each power of 2 the four outputs from the four switches are taken to a Summer which produces an output equal to logical 0 if the sum is even and logical 1 ir it is odd.

The Analyser takes these outputs and decides whether a winning or a losing position is set. In the former case the win output goes low and the 'Your Move' indicator lights. In the latter case the 'My Move' indicator lights and the Analyser initiates a search of the switches to find one which has the highest unpaired power of two present at its output. When this switch is found the appropriate lamp indicator lights.

The last-to-play (L.T.P.) circuitry simply tests for the simultaneous occurrence of no pile containing more than one stick and S5 being set to 'Loser'. In such a situation the control line L goes to logical 1 and the quantity NOT Σ_0 is presented to the Analyser in place of Σ_0 . A little thought will show that this fits the prescription outlined above.

METHOD OF PLAY

The challenger is free to set any initial combination on the switches S1 to S4. The machine should now be switched on and one of the move indicators

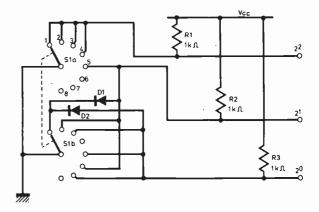


Fig. 2. Wiring detail for one of the decimal to binary switch encoders. Four such arrangements are needed

will immediately light. If 'Your Move' lights the challenger should make some legal Nim move. If the 'My Move' indicator lights then one of the lamps will also be lit and the player should turn the corresponding switch back until the 'Your Move' lamp lights. This represents the machine's move. By continuing in this way, the challenger will invariably lose.

Fig. 2 shows the suggested arrangements of one of the four Encoders S1 to S4, all of which are wired in the same way.

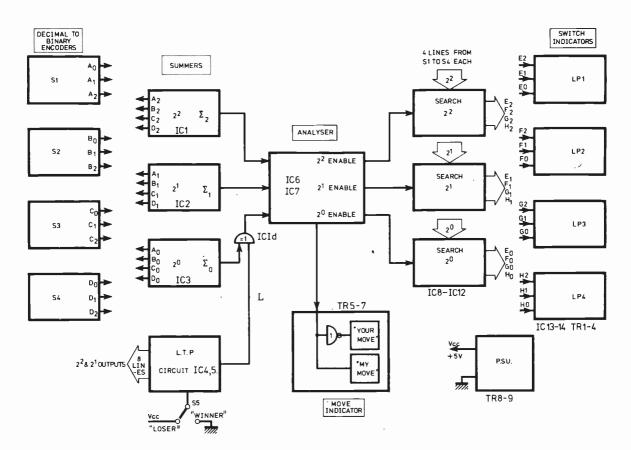


Fig. 1. Block diagram of NIM machine. For detailed wiring of sub-assemblies, see following figures

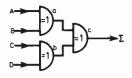


Fig. 3. Three exclusive OR gates make up each Summer. Three such circuits are required

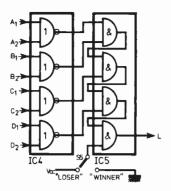
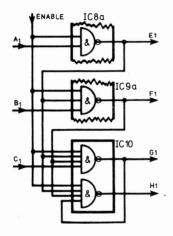
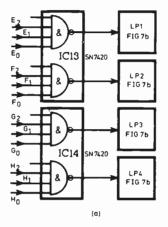


Fig. 4. Arrangement for last-to-play (L.T.P.) circuitry. See Fig. 1 for interwiring detail

Fig. 5. Circuit diagram of Analyser gates





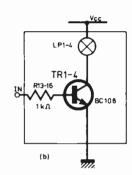


Fig. 6. The Search circuit. Three of these are required

SUMMER

It will be seen from Fig. 3 that the Summer is a very simple circuit block. Each gate employed here is the exclusive or gate, or half adder. Whenever the inputs are identical, the output is 0 but when the inputs differ the output is 1. That this is the function we need may be seen by realising that for our purposes all even numbers are equivalent to 0 and all odd numbers to 1.

Repeated application of the truth table to this arrangement of three gates will show that the final ouput is 1 if the power of 2 is represented an odd number of times and 0 otherwise.

L.T.P. CIRCUITRY

The control line output, L, of the L.T.P. circuit comes from IC5 which is wired as a five input and gate (Fig. 4) needs only to invert Σ_0 when no pile contains more than one stick and the last person to play is deemed the loser. IC4 recognises the former condition by testing the 2^2 and 2^1 outputs from each switch. Only if these are simultaneously zero can that switch be displaying nought or one, and in this case the NOR output is a logical 1.

When the output from each of the NOR gates is 1 and the selector is set to "Loser" a control level 1 is produced which causes IC1d (Fig. 1) to invert the Σ_0 output.

Fig. 7. Switch Indicator circuit (a) output lamp logic (b) lamp driver circuit

ANALYSER

The purpose of the Analyser circuit (Fig. 5) is to find the highest unpaired power of 2. If all powers are paired then the game is set in a winning position, and the Analyser indicates accordingly.

Working from the top we see that if Σ_2 is high then the 2^2 enable line is high also, but IC6b inverts this level which forces both the other enable lines to stay low and forces the 'Win' output high. Similarly if Σ_2 is low but Σ_1 is high the 2^1 enable line is high but the 2^2 and 2° enable lines are low and the 'Win' output is high. Only if Σ_2 , Σ_1 and Σ_0 are all low does the 'Win' output go low and, in this case, all the enable lines are also low. Thus a logical 0 at the 'Win' output corresponds to a winning arrangement.

The Win output controls the Move Indicator directly.

SEARCH CIRCUITS

Suppose that the 2¹ enable line is high (this explanation follows in the same way for both of the other enable lines) then one of the three search circuits in Fig. 1 will be activated. These work in much the same way as the Analyser circuit once they are enabled.

Consider one search circuit in Fig. 6. If the A₁ level is high then E1 goes low and the other gates in the circuit are disabled forcing a high condition

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at each of the other outputs F1, G1 and H1. This will cause lamp LP1 to light. Similarly, if B_1 is high then F1 goes low and the other outputs are forced to stay high so that only LP2 will light.

If A_1 , B_1 and C_1 are all low then H1 goes low and LP4 lights, since the enable line tells us that one of the switches must be presenting a 2^1 contribution.

SWITCH INDICATORS

Each Switch Indicator comprises a NAND gate coupled to a lamp driver and lamp as in Fig. 7. When all three inputs are high the output of the NAND gate is low and the lamp stays off. When a low condition is received from one of the search circuits the output of the NAND gate goes high, and so the lamp lights up. So long as no fault exists, only one lamp can be lit at any one time.

The driver transistors should have a current gain of at least 50—the higher the better—and should be capable of passing up to 100mA. The power dissipation is very small, however, since each transistor is either fully conducting or completely off.

MOVE INDICATOR

The Move Indicator (Fig. 8) consists of two lamp driver circuits coupled by an inverter. The philosophy here is that if it isn't your go, it must be mine!

A discrete circuit is used for the inverter in the prototype, but one of the odd unused gates from the main circuit would work just as well.

The same comments as above apply to the characteristics of the driver transistors used here.

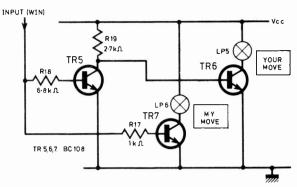


Fig. 8. The Move Indicator consists of two lamp drivers coupled by an inverter

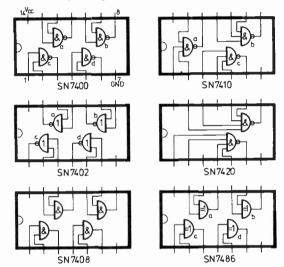


Fig. 9. Pin diagrams for the i.c.s used

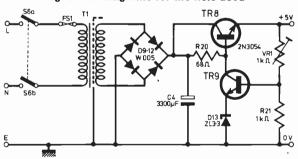


Fig. 10. Suitable power supply for the NIM machine CONSTRUCTION

Providing all interconnecting leads are kept below 10in component layout is not critical. The sub-assemblies given in the figures should be individually assembled and checked electrically before interwiring the complete machine according to Fig. 1.

The pin wiring for the i.c.s is given in Fig. 9 and a suitable power supply in Fig. 10.

The normal i.c. practice of decoupling every 5 to 10 packages with a capacitor of about 0·1 microfarads should be followed.

The use of i.c. holders is not recommended on grounds of cost. The simple gates used in these circuits are available very cheaply, often at prices comparable with the cost of holders.

PATENTS BEVIEW...

PATENTS LIBRARY

Some remarks and reminders on general patent matters bear making. As most readers will understand, the British Patents reported in these columns are selected from the latest batches published (at the rate of around 4,000 a month) by the British Patent Office.

Once a week a new issue of patent specifications is laid out for public inspection in the library attached to the British Patent Office in Southampton Buildings, just off London's Chancery Lane. Similarly, issues are laid out at other libraries round the country. At the same time copies of the specifications are made available for purchase from the Patent Office Sales Branch by any member of the public at 25p each, post free.

At first sight this publishing of protected ideas seems a contradiction in terms. A patent application is made by an inventor who wishes to protect or monopolise his idea for a number of years. While the application is examined by the British Patent Office it is kept rigidly secret from the public; only the title, name and application date are listed in the public card indexes kept in the library. But when the application has been accepted and is published, its content is there for all to read. So what protection is a patent providing for the inventor? And would it not have been hetter for them simply to keep the details of his invention secret and not involve himself in the expense of securing a patent (likely to be at least £100 if handled by a Patent Agent)?

PATENT CONTENTS

A patent specification is worded in a language which many people regard as an incomprehensible mixture of legal and technical jargon. After a legally worded preamble, every specification launches into a general description of the invention and then moves on to a specific description (usually with drawings) which explains in detail how one or more practical embodiments of the invention actually work.

Every patent specification ends with a set of claims which are the

most important part when monopoly dinfringement are considered. The claims define the scope of monopoly which the Patent Office has agreed to allow the inventor, having carried out searches through

previously published patents.

If a claim is too broad, it may "catch" a large number of "infringing" articles, but it may also embrace articles that are known to be old. So the claim is in practice invalid. And anyone concerned and believing that the claims of a freshly published patent are obviously invalid can either ignore them or

legally oppose them at a Patent Office Hearing.

ADVANTAGES OF PUBLISHING

Publishing patents also serves to enhance the richness of human knowledge and once a patent has expired (even if the full £300 worth of annual renewal fees are paid, a British patent can only last 16 years) all that it discloses and claims becomes part of the public domain. The basic reasoning is that the inventor is granted a limited monopoly in return for disclosing his idea rather than keeping it secret. Incidentally, laying open the details of an invention to the public also serves as something of an advertisement for the owner of the patent, who may be approached by manufacturers who wish to make what he claims under a licence.

It is a basic canon of patent law throughout the world that (subject to a very few exceptions) no one can patent an invention once it has been disclosed to the public. Thus, usually, once an invention has been published (e.g. in a book or magazine, or offered for sale) no one, the inventor included, can subsequently patent it. But it is perfectly safe for an inventor to lodge an application (even if it is accompanied only by a Provisional Specification, which simply lies fallow at the Patent Office for a year) and then immediately publish his invention. Making the patent application generates a priority date which safequards subsequent disclosure.

Where details of a circuit or construction are given in a magazine with d-i-y slant, e.g. Practical Electronics or Everyday Electronics it would not normally be possible for anyone to apply for a patent,

for anything disclosed, after publication. However, if a contributor had already applied for a patent before the appearance of the article describing his design, then the appearance of that article would not in itself invalidate any patent subsequently granted.

LEGAL ACTION

Because the long arm of patent law moves slowly, there is often a gap of several years between original application and final acceptance and grant. Thus it is perfectly possible that a circuit or design published in a magazine such as this could already be the subject of a pending patent application which will not come into legal force until months or even years after the appearance of the relevant magazine issue.

Usually no action for infringement can ever be taken on a pending patent application (remember, they are kept secret so it would be grossly unreasonable if such action could be taken) this could under certain circumstances mean that designs to be found in some back numbers are only now protected

by patents.

In practice these circumstances have never presented problems and are unlikely ever to do so, largely because the law is not often an ass. For instance, it is so unikely as to be inconceivable that the owner of a patent would ever be in a position to know who was making a "one-off" model of his design in the privacy of his own home or garden shed.

But anyone considering an expensive production run of any construction should always try and discover whether or not patent rights (perhaps owned by some completely different third party) are involved before embarking on such a project. For this reason it would seem in the interests of everyone if contributors with d-i-y projects included a reference by number to any relevant patents or patent applications.

Finally, readers believing that they have a patent problem on their hands should seek advice from a Patent Agent; no one on Practical Electronics can enter into correspondence on such matters.

Copies of Patents can be obtained from the Patent Office Sales, St. Mary Cray, Orpington, Kent. Price 25p each

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*Constant and algebraic logic combine to act as a limited memory, allowing complex calculations on a calculator costing less than £17.

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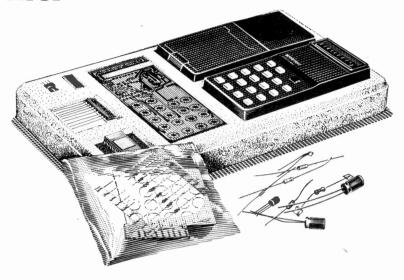
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RECTION FROM OUR POSTBAG

Good olde days!

For over twenty years, I have been an inveterate constructor. In my early days they were the designs of people more clever than myself. Then, with increasing knowledge and experience. I began to try my hand at designing my "own" circuits.

An essential part of my apprenticeship was acquiring the expertise essential if one was to become a skilful "chassis basher", out of aluminium, mild steel or tin plate. A chassis was an essential prerequisite for use with valves, which were, for the benefit of those who haven't seen one, made of glass with bits of metal inside them. The valves usually glowed dull red (that's the filament, son), sometimes got very hot, and sometimes "bit" hard if one got careless. I ended up in hospital once when I got careless whilst working on high power transmitters.

A well laid out, and carefully constructed chassis, imbued its constructor with a feeling of euphoria, whilst he sniffed at the delicate aroma of hot valves, warm Bakelite and warm transformer impregnant, occasionally enlivened by over heated resistors.—What joy! What

contentment!

Then along came the transistor. I had to discard most of my painfully acquired wisdom and start learning again. Early results were most discouraging, but I persevered, and, slowly, success came my way. But some of the pleasure had gone. A tag board here, a turret tag there. The resulting rat's nest on an s.r.b.p. board possibly worked, but it looked ghastly and meant that one less skill to be exercised.

To cap it all, along came i.c.s, and, again, less skill was required on the part of the constructor. A few i.c.s, and sundry components mounted on a bit of "holey" s.r.b.p. board (Veroboard) and there it was. It may have worked, but did it really take much "constructing"?—

I personally don't think so.

Over the period of time since I first entered the world of electronics, the individual constructor has had to exercise less and less skill, more and more is done for him. In this world of "instant" instants, who has the time to learn a new set of skills to really build or construct something different, something unique?

It is a chastening thought, and, I think, a saddening one, to reflect on the electronics constructors lot in, say, ten years' time. You want to "build" a radio, son? You'd like to build a 50W stereo amplifier? Here are three "cubular modules", press the top of that one, there's the radio you built. Plug the other two into the mains, plug in your sound source and there's the stereo amplifier you built.

Where's the fun? Where's the pleasure of actually constructing something with your own hands? Fun? Pleasure? I for one, mourn the passing of the early transistor days.—Those were the days.

H. T. Kitchen, Bulkington, Warks.

Fog warning system

Sir,—With the perennial concern for Motorway safety I feel you should know about a fog warning system I have devised after a considerable period of experimentation.

In principle it monitors the whole of the Motorway, is automatically triggered by fog or smoke and provides advance warning to motorists 300m prior to the affected area.

The basic system is as follows (See Fig. 1). A light beam is directed along the Motorway verge in the direction of oncoming traffic to an optical receiver 300m away. While the beam is maintained a relay controls another signal lamp directed to a receiver a further 300m along the Motorway verge and so on. Such a chain is optically possible since Motorways are practically

straight and level and there are no obstructions along the verges.

If the beam is interrupted by fog, heavy rain, snow or smoke the receiver causes the relay to change over and transfer power from the following signal lamp to multivibrator controlled warning lamps. The net result is that all lamps in the chain are cancelled and warning lamps are left to flash warning of danger.

The optical receiver consists of a long collimator tube, matt black inside, which directs the light beam from the lamp to a 4½in concave mirror and light dependent resistor which almost forms the focus. Similarly the transmitter has a 4½in concave mirror fed from a 10V, 1A bulb. This supply is obtained from a tapped car battery (older kind) which also sources two groups of six 3.5V, 0.3A bulbs in a series parallel arrangement making up the collector loads to a multivibrator circuit.

A modification to the system is the inclusion of a lamp pulser (multivibrator). This permits the system to be used over long distances. If, say, fog were were only affecting a section of the Motorway the closing of the relay would operate the warning lamps and the pulser. For that transmitter path where there was no fog the pulser would cause the lamp to flash, this would restore the "all-clear" to the other units along the Motorway.

By carefully timing the pulses and adding a suitable delay to each receiver it is possbile to have two fog affected units showing a warning and the rest of the chain show-

ing "all-clear."

From the circuits devised, field trials have been encouraging. Some further experiments have been conducted using modulated light, but it is a matter for further study as to which of these ideas is the most suitable, but the writer believes that it should not be beyond the abilities of, say, students working for A level physics to devise a practical solution, working on these lines, to a national problem.

G. Cozens, Bartley, Southampton

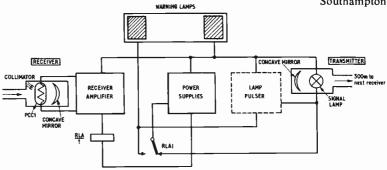


Fig. 1. Basic receiver/transmitter unit for controlling fog warning lamps on the Motorway

Psycho-sensitive semiconductors?

Sir—With reference to Mr. Brian Baily's article in P.E. April '74 about ESP I should like to propose the following explanation of the results he obtained from his experiments on plants. I wish to point out that this is merely a hypothesis and that I have no evidence other than that which I state:

In one of the programmes in the recent BBC T.V. series "Young Scientist of the Year!" the results of a group of boys' experiments on growing plants in magnetic fields, were presented. They found that plants grown in a magnetic field bent and grew along the lines of force. Could this not be something to do with the fact that there is an atom of iron at the core of the chlorophyll molecule, which may render it magnetically sensitive?

It is possible that the molecules align themselves along the lines of force, presenting a coherent chain of electrically resistive molecules, whereas normally they would be in an amorphous distribution. This migration of the vital chlorophyll towards one point would explain the tropic response of the plants in the magnetic fields.

Is it not possible then, that if a large, strong field is enough to cause the plant to change its direction of growth, the weak field surrounding an electrically active brain would be sufficient to align one or two

chains of molecules and thus cause a change in the cells' resistance?

The shrimps' nervous system on contact with the boiling water would radiate a field of far greater strength than normal, this would be selected by the plant and the appropriate change in resistance would occur. I propose that a concentrating mind, regardless of the subject of that concentration, produces a field strong enough for the plant to detect.

Assuming this theory to be correct, there must be other chemicals possessing the same property, in which case it is not hard to imagine, in the not-too-distant future, the invention of a psycho-sensitive semiconductor device. This would be a boon for handicapped people who would be able to turn on equipment merely by thinking about it, no physical help required.

P. Watson, Bedfordshire.

Hot line

Sir—A short while ago, my wife and I decided to change the position of the furniture in our lounge. The arrangement we chose meant that the colour television would not plug directly into the socket, so I decided to use an extension cable. The only one which was available was a 150ft cable which I use with the electric lawn mower.

Being the cautious type I checked that the television would not overload the cable and as the cable was rated at 5A and the television consumption 245W, which is just over 1A at 240V, I decided that it would be safe. As you can imagine, 150ft of cable across a lounge leaves quite a lot to spare so I coiled it neatly and put it behind the television.

I offered a neighbour of mine the use of my lawn mower as he was having trouble coping with a thick growth. As the cable on the machine was not long enough I also offered the extension cable, this was when I found to my horror that the neatly coiled cable had welded itself together, it had apparently acted as a heating coil with IA being sufficient to create enough heat to ruin all but 15ft of the cable. During this time the cable has not short circuited or failed but it has left a nasty burn mark on the carpet.

I had wondered why the cat had taken to sleeping under the television, she usually finds the warmest position to curl up.

R. F. Burgin, Witham, Essex.

According to l.E.E. regulations current ratings for cable are based on a fixed amblent air temperature. Any heat arising from the cable caused through coiling must derate this figure. The combination of heat generated in a long cable magnified by coiling and the cat probably contributed to the thermoplastic insulation deteriorating.—Ed

SOUND SYNTHESIS FOR THE AMATEUR



Due to the unprecedented number of letters we have received from readers stating their disappointment at not being able to attend the P.E. Audio Fair lectures, and more recently the lecture given at Essex University, we have decided to make available (at a small fee) to Clubs, Schools and Universities a taped recording and colour slides of the lecture.

Entitled "SOUND SYNTHESIS FOR THE AMATEUR" the tape runs for approximately 45 minutes at $7\frac{1}{2}$ l.p.s. and was recorded on a Philips N4450 domestic recorder. The colour slides contain waveforms, block diagrams and circuits on the various sections of the P.E. Synthesiser. Only two sets of tapes and slides are available and requests will be dealt with in strict rotation of receipt.

It is regretted that this service can only be offered to CLUBS, SCHOOLS, and UNIVERSITIES. The charge is £1 to cover handling and postal charges. A refundable deposit of £5, to cover any damage to tapes or slides, is also required.

Requests for the tape and slides (with date of proposed use) should be addressed to:

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26304	1.15	BF197 BF861	0-15	NKT216 NKT217	0-40 0-45	OC139 OC140	0.40	7498	0.75
26501 26703	0-75 1-00	BF898	0-25 0-25	NKT218	1.18	OC140 OC141	0-65 0-80	7494	0.85
AA129	0.20	BFX12	0.80	NKT218 NKT219 NKT222	0.88	OC169	0-20	7495 7496	0-85 1-00
AAZ12	0.75	BFX13	0-25	NKT222	0.80	OC170	0-25	7497	4-82
AAZ13 AO107	0.10	BFX29 BFX30	0-28 0-28	NKT224 NKT251	0-25 0-24	OC171 OC200	0-80 0-55	74100	2-16
AO126	0-25	BFX 85	0-98	NKT271	0-20	OC201	0-80	74107 74110	0-51 0-57
AC127	0.25	BFX63	0-50	NKT272	0.20	OC202 OC203	0.90	74111	0-86
AC128	0-20	BFX84 BFX85	0-25	NKT273 NKT274	0-20 0-20	OC203 OC204	0-55 0-65	74118	1.00
AC187 AC188	0-20	BFX86	0.25	NKT275	0.25	OC205	1.00	74119 74121	1.98
ACY17	0-85	BFX87	0.25	NKT277 NKT278	0-20 0-25	OC205 OC206	1.10	74122	0.80
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ACY21	0-22	BFY17 BFY18	0-40 0-45	NKT403 NKT404	0.70	OCP71	1.00	74150	2-80
ACY22 ACY27	0-16 0-25	BFY19	0-55	NKT678	0.80	ORP12 ORP60	0-55	74151	1-15
ACY28	0-25	BFY19 BFY24	0-45	NKT678 NKT713	0-80	ORP61	0-48	74154 74155	2.80 1.15
ACT39	0-65	BFY44 BFY50	1.00	NKT773 NKT777	0-25 0-38	8X68	0-20	74155	1-15
ACY40	0-22	BFY51	0-20	078B	0-38	8X631 8X635	0-45	74157	1.09
ACY41 ACY44	0-88	BFY52	0.20	O A 5	0.60	8X640	0.75	74170	2-88
AD140	0.50	BFY53	0-17	OA6	0.18	8X641	0.75	74174	1.80
AD149 AD161	0-50	BFY64 BFY90	0-45 0-75	OA47 OA70	0-08 0-10	8X642 8X644	0-60 0-85	74175	1.29
AD161	0-88	B8X27	0-50	OA71	0-20	8X 645	0.85	74176	1-44
AF106	0.80	BSX60	0-98	OA73	0.15	TIC44	0-29	74190 74191	2-80 2-80
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60 0 Ref. No. 07 149 150 151 152 153 154 155 156 158 7-0 × 6-0 × 6-0 9-9 × 7-7 × 8-6 9-9 × 8-9 × 8-6 12-1 × 9-3 × 10-2 12-1 × 11-8 × 10-2 14-0 × 10-8 × 11-8 17-2 × 14-0 × 14-0 17-2 × 16-6 × 14-0 21-6 × 15-3 × 8-1 8 12 8 0 12 0 8 0 0 P & P £ 1:34 7 2:64 3 3:18 6:19 9:20 16:71 24:19 31:57 39:17 AUTO TRANSFORMERS Size cm. Auto Taps VA Weight (Watts) Ib oz 20 | 0 | 0 | 75 | 2 | 4 | 150 | 3 | 4 | 300 | 6 | 4 | 1500 | 12 | 8 | 1 | 1500 | 30 | 4 | 1 | 2000 | 32 | 0 | 1 | 3000 | 40 | 0 | 2 Ref. No. 113 64 66 67 84 93 95 73 5-8 × 5-1 × 4-5 0-115-210-240 7-0 × 6-7 × 6-1 0-115-210-240 8-9 × 7-7 × 7 0-115-200-220-240 9-9 × 9-6 × 8-6 12-1 × 11-2 × 10-2 14-0 × 13-4 × 14-3 14-0 × 15-9 × 14-3 17-2 × 16-6 × 14-0 21-6 × 13-4 × 18-1 CASED AUTO TRANSFORMERS
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17 16 8 8 12 12-1 × 9-9 × 10-2 0-12V at 52 × 2 5-61
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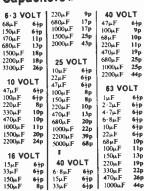
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3-WAY STEREO HEAD-PHONE JUNCTION BOX

2-WAY CROSSOVER NETWORK

K 4007. 80 ohms Imp. Insertion loss 3d B £1.21

SELECTED RESISTORS

CAR STEREO SPEAKERS Angled) £3.85 per pair

BI-PAK

CATALOGUE AND LISTS

Send S.A.E. and 10p.



No.	Length	11	idth	He	ight	Price
BVI	8"	×	51"	×	2"	90p
BV2	11"	×	6*	×	3"	£1.20

ALL	IMIN	HUH	BO	KES		
BAI	51"	X	24"	×	13"	4:
BA2	4"	×	4"	×	14"	- 4
BA3	4"	×	23-	×	11"	4
BA4	51"	×	4"	×	13"	4
BA5	4"	×	21"	×	2"	- 4
BA6	37	×	2"	×	1"	3
BA7	7"	×	ō"	×	21-	6
BAS	8"	×	6"	×	3"	8
RAG	65.	~	4"	×	->"	5

BIB HI-FI ACCESSORIES

De Luxe Groov-Kleen Model 42 £1:84

Chrome Finish Model 60 £1:50



Ref. 36A. Record/Stylus Cleaning Kit 28p Ref. 43. Record Care Kit £2-35 Ref. 31. Cassette Head Cleaner 54p

Ref. 32 Tane editing Kit \$1.54 Model 9 Wire Stripper/Cutter 83r Ref. P. Hi-Fi Cleaner 31p Ref. 32A. Stylus Balance £1.36 Ref. J. Tape Head Cleaning Kit 51p Ref 56 Hi-Fi Stereo Hints & Tips 32p

ANTEX SOLDERING IRONS

X25, 25 watt £1.93 CCN 240.(15 watt £2-15 Model G. 18 watt £2-15 SK2. Soldering Kit \$2.86 STANDS: STI #1-21, ST2 77p SOLDER: 188WG Multicore 7oz 82p 228WG 7oz 82p. 188WG 22ft 28p 228WG Tube 22p

ANTEX BITS and ELEMENTS

3its		
	For model CN240 32"	
102	For model CN240 A	1
100	For model CCN 240 32"	1
10 i	For model CCN240 1"	1
102	For model CCN240 1"	ALL
020	For model G240 🔥 "	38p
021	For model G240 §"	each
022	For model G240 #"	
50	For model X25 🖧 "	1
51	For model X25 #"	1

52 For model X25 #" ELEMENTS

ECN 240 £1-16 EG 240 £1-16

ECCN 240 £1-32 EX 25 21-16

ANTEX HEAT SINKS 10p

VAT included in all prices. Please add 10p P. & P. (U.K. only). Overseas orders—please add extra for postage.

NEW COMPONENT PAK RARGAINS

No. Qty. Description C1 250 Resistors mixed values a count by weight C2 200 Capacitors mixed values a	Price pprox. 0.55
count by weight	0.55
(5) 200 Capacitors mixed values a	DEPOY
count by weight	0.55
C3 50 Precision Resistors 0.1%, mixed values	0.01% 0.55
C4 75 4th W Resistors mixed provalues	eferred 0.55
(5 5 Pieces assorted Ferrite Rods	0.55
C6 2 Tuning Gangs, MW/LW VHI	
C7 7 Pack Wire 50 metres as	norted 0.55
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C9 3 Micro Switches	0.55
C10 15 Assorted Pots & Pre-Sets	0.55
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C19 3 Relays 6-24V Operating	0.55
C20 4 Sheets Copper Laminate 10" × 7"	approx. 0.55

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18 BALDOCK ST., WARE, Herts. (A10) Open Monday-Sat. 9-5.30 p.m. Tel. 61593

Ref. 34. Cassette Case £1.27

PLUGS	AND SOCKETS	
SOCKETS		
PS 35 DIN	2 Pin (Speaker)	0.06
PS 36 DIN	3 Pin	0.10
PS 37 D1N	5 Pin 180°	0.10
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PS 39	Jack 2-5mm Switched	0.09
P8 40	Jack 3-5mm Switched	0.10
P8 41	Jack 1" Switched	0.17
PS 42	Jack Stereo Switched	0.26
PS 43	Phono Single	0.06
PS 44	Phono Double	0.10
PS 45	Car Aerial	0.09
P8 46	Co-Axial Surface	0.08

Co-Axiai Finsh

0.14

INLI	NE SOCKETS	
P8 21	D.I.N. 2 Pin (Speaker	0.13
PS 22	D.I.N. 3 Pln	0.17
P8 23	D.I.N. 5 Pin 180°	0.17
PS 24	D.I.N. 5 Pin 240°	0-17
PS 25	Jack 2-5mm Plastic	0.10
PS 26	Jack 3.5mm Plastic	0.18
PS 27	Jack 1" Plastic	0.24
PS 28	Jack 1" Screened	0.28
PS 29	Jack Stereo Plastic	0.25
PS 30	Jack Stereo Screened	0.88
PS 31	Phono Screened	0.14
PS 32	Car Aerial	0.18
PS 33	Co-Axial	0.1

DILLGS

PS 16 Co-Axial

PS 47

PS	1	D.I.N. 2 Pin (Speaker)	0.11
P8	2	D.1.N. 3 Pin	0.12
Р8	3	D.I.N. 4 Pin	0.15
P8	4	D.I.N. 5 Pin 180°	0.14
P8	5		0.15
P8		D.I.N. 6 Pin	0.15
P8			0.15
PS		Jack 2-5mm Screened	0.10
PB	-	Jack 3-5mm Plastic	0.09
P8		Jack 3.5mm Screened	0.12
P8		Jack i" Plastic	0.13
PB		Jack 1" Screened	0.18
PS			0.29
P8			0.06
100		Control of the state of the sta	0.15

CP	1	Single Lapped Screen	0.06
CP	-2		0.08
CP	3	Stereo Screened	0.08
CP	4	Four Core Common Screen	0.23
PC	5	Four Core Individually Screened	0.80
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C

0.10

ARTRIDGES	
COS GP91-18C. 200mV at 1-2cms/sec	£1-16
COS GP93-1, 280mV at 1cm/sec	£1.65
COS GP96-1, 100mV at 1cm/sec	£2.65
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The E12 Range of Carbon Film Resistors, 1 watt available in PAKS of 50 pieces, assorted into the following groups:— RI 50 Mixed 100 ohms-820 ohms R2 50 Mixed 1K ohms-8-2K ohms 40p R3 50 Mixed 10K ohms 82K ohms 40p R4 50 Mixed 100K ohms-1 Meg. ohms 40p

BI-PAK SUPERIOR QUALITY LOW-NOISE CASSETTES

THESE ARE UNBEATABLE PRICES-LESS THAN 10 EACH INCL. V.A.T.

C60, 32p; C90, 41p; C120, 52p

-the lowest prices

BI-PAK QUALITY COMES TO AUDIO!

AL10/AL20/AL30 **AUDIO AMPLIFIER MODULES**



The AL10, AL20 and AL30 units are similar in their appearance and in their general specification. However, careful selection of the plastic power devices has resulted in a range of output powers from 3 to 10 watts R.M.S.

The versatility of their design makes them ideal for use in record players, tape recorders, stereo amplifiers and cassette and cartridge tape players in the car and at home.

Parameter	Conditions	Performance
HARMONIC DISTORTION	Po = 3 WATTS f = 1KHz	0.25%
LOAD IMPEDANCE	_	8-16 Ω
INPUT IMPEDANCE	f = 1 KHz	100 kΩ
FREQUENCY RESPONSE -3dB	Po = 2 WATTS	50 Hz-25KH
BENSITIVITY for RATED O/P	$V_8 = 25V$. $R1 = 8\Omega$ $f = 1KHz$	75mV. RM8
DIMENSIONS	-	3" × 21" = 1"

The above table relates to the AL10, AL20 and AL30 modules. The following table outlines the differences in their working conditions.

Parameter	AL10	AL20	AL30	
Maximum Supply Voltage	25	30	30	
Power out for 2% T.H.D. (RL = 8Ω f = 1KHz)	3 watts RMS Min.	5 watts RMS Min.	10 watts RMS Min.	

AUDIC AMPLIFIER

MODULES AL 10. 3 watts AL 20. 5 watts AL 30. 19 watts £2.19 £2.59 £3.01

POWER SUPPLIES

PS 12. (Use with AL10 & AL20) 8
3PM 80. (Use with also AL30 & AL50) 88n FRONT PANELS PA 12 with Knobs

PRE-AMPLIFIERS

PA 12. (Use with AL10 & AL20) £4-35 PA 100. (Use with AL30 & AL50) £13-15

TRANSFORMERS

T461 (Use with AL10) 21.38 P & P 15p T538 (Use with AL20) £1.93 P & P 15p BMT80 (Use with AL30 & AL50) £2.15

PA12 PRE-AMPLIFIER SPECIFICATION

The PA12 pre-amplifier has been designed to natch into Frequency response—
20Hz-50KHz(-3dD) nost budget stereo systems. It is compatible with the AL 10, AL 20 and AL 30 audio power amplifiers and it an be supplied from their associated power supplies. There are rwo stereo inputs, one has been designed for use with "Ceramic cartridges while the auxiliary input will uit most 'Hagnetic cartridges. Pull details are given in he specification table. The four controls are, from left to ight: Volume and on/off switch, balance, bass and treble. Size 152mm × 84mm × 35mm.

20H2-30KH2 (-3dD)
Bass control—

± 12dB at 60Hz
Treble control—

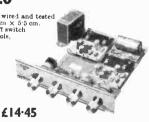
± 14dB at 14KHz *Input 1. Impedance
1 Meg. olun
Sensitivity 300mV
finput 2. Impedance
30 K ohms
Sensitivity 4mV

LOOK FOR OUR SEMICONDUCTOR ADVERTISEMENTS IN Practical Wireless and RADIO CONSTRUCTOR

ALL PRICES INCLUDE VAT

The STEREO 20

The "Stere» 20" amplifier is mounted, ready wired and tested in a one-piece chassis measuring 20 cm \times 14 cm \times 5.5 cm. This compact unit comes complete with on/off awitch olume control, balance, bass and treble controls, olume control, balance, base and treble contransformer, Power supply and Power amps. ttractively printed front panel and matching control knobs. The "Stereo 20" has been lesigned to fit into most turntable plinthal rithout interfering with the mechanism or, lternatively, into a separate cabinet, butput power 20w peak. Input 1 (Oer.) 00mV into 1M. Freq. res. 25Hz-25kHz. ppt 2 (Aux.) 4mV into 30K. Harmonic listortion. Base control ± 12dB at 60Hz ypically 0.25% at 1 wstt. Treble con. ± 14dB at 14kHz.



NOW WE GIVE YOU 50w PEAK (25w R.M.S.) PLUS THERMAL PROTECTION I The NEW AL60 Hi-Fi Audio Amplifier FOR ONLY £3.95

- Max Heat Sink temp 90°C.
- Frequency Response 20Hz to 100KHz
- 0.1% Distortion
- Distortion better than 1% at 1KHz
- Supply voltage 10-35 volts
- Thermal Feedback
- Latest Design Improvements
- Load 3, 4, 8 or 16 ohms
- Signal to noise ratio 80dB
- Overall size 63mm × 105mm × 13mm

Especially designed to a strict specification. Only the finest components have been used and the latest solid state circuitry incorporated in this powerful little amplifier which should satisfy the most critical A.F. enthusiast.

FULLY BUILT - TESTED and GUARANTEED

STABILISED POWER **MODULE SPM80**

AP80 is especially designed to power 2 of the AL50 Amplifiers, up to 15 watt (r.m.s.) per channel simultaneously. This module embodies the latest components and circuit techniques incorporating complete short circuit protection. With the addition of the Mains Transem MT80, the unit will provide outputs of up to 1-5 amps at 35 volts. Size: 63mm × 105mm × 30mm.

These units enable you to build Autilo Systems of the highest quality at a hitherto unobtainable price. Also ideal for many other applications including:—Disco Systems, Public Address, Intercom Units, etc. Handbook available 10p PRICE £3*25

TRANSFORMER BMT80 £2:15 p. & p. 28p

STEREO PRE-AMPLIFIER TYPE PA100

Built to a specification and NOT a price, and yet still the greatest value on the market, the PA100 stereo pre-amplifier has been conceived from the latest circuit techniques. Designed for use with the AL60 power amplifier system, this quality made unit incurporates no less than eight silicon planar transistors, two of these are specially selected low noise NPN devices for use in the input stages. Three switched stereo inputs, and rumble and scratch filters are features of the PA100, which also has a STEREO/MONO switch, volume, balance and continuously variable has and treble controls.

bass and treble controls.



SPECIFICATION

Bass Control Treble Control
Filters: Rumble (High Pass)
Scratch (Low Pass)
Signal/Noise Ratio

Input overload Supply naiona

 SPECIPICATION

 Frequency Response
 20Hz--20KHz ± 1dB

 Harmonic Distortion
 better than 0·1%

 Inputs 1. Tape Head
 1·25 mV into 50K Ω

 2. Radio, Tuner
 35 mV into 50K Ω

 3. Magnetic P.U.
 1·5 mV into 50K Ω

 All input voltages are for an output of 250mV. Tape and P.U. inputs equalised to RIAA curve within ± 1dB. from 20Hz to 20KHz.

 iass Control
 ± 15dB at 20Hz

 reble Control
 ± 15dB at 20Hz

 ilters: Rumble (High Pass)
 100Hz

 Scratch (Low Pass)
 8KHz

 ignal/Noise Ratio
 better than -05dB

 pupt overload
 + 26dB

+ 26dB + 35 volts at 20mA - 90mm × 35mm

ONLY £13:15

SPECIAL COMPLETE KIT COMPRISING 2 AL50's, 1 SPM80, 1 BMT80 & 1 PA100 ONLY £25'30 FREE p. £ p.

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sound source from I to 100 watts produces a psychodelic light display of up to 1000 watts. Complete with a sensitive level control the unit is fused and cannot harm your amplifier.

A Bargain at £7.50 plus 10p.

"ORESCENT" BUBBLE LIGHT SHOW



A new and exciting feature for the professional disc jockey or to give the private party an electric atmosphere, a projected kaleido-

atmosphere, a projected kalelidoscope of colour.
Specification—Projector: 100W, convection cooled, at 30th the projected image = 16ft; Motor: 1 rev per 2 min. Liquid Wheel: din diameter multicolour. The Motor is fitted to the Projector and can only be purchased as a single unit. The Liquid Wheel, however, is our very popular standard model & may be purchased separately. A bargain—Projector with Motor ready for instantuse, £15; 6in Liquid Wheel, £8=220 + 75p carr.



TRI-VOLT BATTERY ELIMINATOR

Enables you to work Transistor

your Transistor Radio, Amplifier or Cassette, etc., from the a.c. mains through this compact the a.c. mains through this compact Eliminator. Just by moving a plug you can select the voltage you require, 6, 74 or 9 volt. This means all your transistor power pack applications can be handled by this one unit. Approx. size 24in × 24in × 34in. Our Price 52.75 plus 10p P. & P. Same model suitably wired for the Philips Cassette 28 plus 10p P. & P.

7in ×4in LOUDSPEAKER



A top quality speaker ideal where small size is import-

small size is important. Manufactured by E. M. I. for a well-known hi-fi set 4in. Impedance: 8 ohms. Flux: 38,000. Max. Free range: 90Hz to 12kHz. Power handling: 5W. Unbeatable Frice: 21-60. Free postage on this item

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If you construct you should own one. Send 20p inc. carriage.



UK65	Transistor Tester	21.66
UK92	Telephone Amp	£8-26
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UK 130	Mono Control Unit	24-15
UK 145	Amp-1-5W	£3.81
UK 165	RIAA Equalised Stereo Amp	£5.30
UK 195	Mini-Amp-2W	£3-66
UK 220	Signal Injector	£2-65
UK230	AM-FM Aer. Amp	£3-29
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TRI-VOLT CAR SUPPLY
Enables you to work your Transistor Radio,
Amplifier or Cassette, etc. from the 12 voit car
supply. Positive or negative earth. Approx.
size = 2m × 3 im × 1 im. This converter
supplies 6, 7 io 7 voits and is transistor regulated.
A real money saving device for \$2.50, 10p P. & P.

"C 300" DISCO CONTROL PACK

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A control Unit which when connected to twin decks
make a disco of professional quality. We supply a
smart front panel which incorporates controls,
switch and input sockets. The control module, I.C.
construction incorporating mixing, pre-amp and
headphone listening amplifier. The power pack
enables this unit to work from the standard mains.
Ipputs include Mic., Tape/Cassette and Twin
Decks. * Controls include Mic., Tape, Each Deck,
Mono, £14. Stereo, £17 plus 20p carr.

KILOWATTS PSYCHEDELIC LIGHT CONTROL UNIT



Three Channel: Bass—Middle—Treble. Each channel has its own sensitivity control. Just connect the input of this unit to the loudspeaker terminals of an amplifier, and connect three 250V up to 1000W lamps to the output terminals of the unit, and you produce a fascinating sound-light display. (All guaranteed) £18.50 plus 38p P. & P.

LOW VOLTAGE AMPLIFIER

5 transistor amplifier complete with volume control, is suitable for 9V d.c. and a.c. supplies. Will give about 1W at 8 ohm

With high IMP input this ampliwork as a record baby alarm, etc.,

£1.75 plus 13p P. & P

200/250V MAINS RELAY



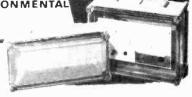
MINI LOUDSPEAKERS 2½in 80 ohm, 50p; 2½in 40 ohm, 50p. Please include 5p P. & P. on each L.S.

VAT

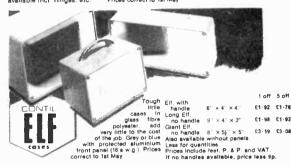
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A smart miniature case in tough, rigid, high-gloss black A.B.S. Front panels in either aluminium or white PVC/Steel Built-in slots for PC cards, dividers, or

PC boards can be supported on clips from internal pillars. Exceller encapsulation boxes

1 off 10 off M2 65mm × 100mm × 45p 50mm2\(\frac{1}{2}\)" \times 3\(\frac{1}{2}\)" \times 2\(\frac{1}{2}\)" \times 2\(\frac{1}{2}\)" \times 3\(\frac{1}{2}\)" \times 2\(\frac{1}{2}\)" \times 5\(\frac{1}{2}\)" \times 2\(\frac{1}{2}\)" 51p 67n 580



Prices include P. & P. and VAT. Less for quantities. Also available without panel and screws. Minimum order £1.
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W. H. D. 1 off W. H. D. 1 off A 4.5" 3" 6.5" 62.90 M 4.5" 3" 13" 12" 52.44 C 4.5" 10" 6.5" 62.44 N 4.5" 7" 13" 52.44 C 4.5" 10" 6.5" 62.41 P 9 3" 13" 13" 52.44 E 9 3" 10" 6.5" 62.41 P 9 3" 13" 13" 52.44 E 9 3" 10" 6.5" 62.41 P 9 3" 13" 13" 52.44 E 9 3" 10" 6.5" 62.44 S 13" 10" 13" 62.41 F 9 3" 10" 13" 10" 13" 62.41 F 9 3" 10" 13" 10" 13" 10" 13" 10" 13" 10" 13" 10" 13" 10" 13" 10" 13"

The design of these cases permits the instrument to be built or serviced within their external panels. 48 shapes. Low cost. Blue PVC/Steel with white P.V.C. aluminium panels.

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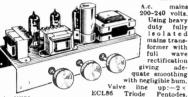


A superb solid state audio amplifier. Brand new components throughout. 5 silicon transistors transistors plus 2 power output transis tors in push-pull. Full wave rectification. wave recuncation. Output approx. 13W r.m.s. into 8 ohm.

Frequency response 12Hz-30kHz + 3db. Fully integrated preamplifier stage with separate Volume, Bass boost and Treble cut controls. Suitable for 8-15 ohm speakers. Input for ceramic or crystal cartridge. Sensitivity approx. 40mV for full output. Supplied ready built and tested, with knobs, escutcheon panel, input and output plugs. Overall size 3in high x 6in wide x 7iin deep. A.C. 220/250V.

PRICE £12.00 P. & P.

DE LUXE STEREO AMPLIFIER



Valve I line up: 2-2 With negligible hum. Valve I line up: 2-2 ECL86

FOL86 Triode Pentodex potentiometers are provided for baw and treble control, giving bases and treble boost and cut. A dual volume control is used. Balance of the left and right hand channels can be adjusted by note that the rear of the chassis. Input sensitivity is approximately 30m for full peak output of 4 watts per channel (8 watts mono) in of any of the chasses and the sensitivity of the chasses and the sensitivity of the proximately 30m carefully calculated circuit, allows high volume levels to be used with negligible distortion. Supplied complete with knobs, chasses size II in. w. x din. x. Overall height including valves 5 in. Ready built and tested to a high standard. Price £10.40, P. & P. 50p.

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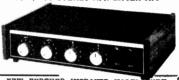
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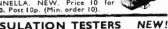
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OUR PRICE £5.95

HUNKI Model 720X VOM

A versatile, accurate measuring instrument. 20,000 opv. 0/5/25/100/500/1000V DC. 0/10, 50/250/1000V AC. 0–50uA/250mA. 0–20k/2 Megohms.



MODEL PL436 muuel rL43 20,000 opv DC. 8000 opv AC. Mirror scale -6/3/12/30/120/ 600V DC. 3/30/ 120/600V DC. 50/600μA/60/

10/100K/1 Meg/10 Meg Ohm.

OUR PRICE £6.97 P&P 15p

U4323 MULTIMETER

20,000 opv. Simple unit with audio/IF cacillator. Suitable moritator. Suitable far general receiver toning. Ranges: 0.5/2.5/10/50/250/ 500/1000V DC. 2.5/10/15/250/500/1000V AC. 0.05/

plants with test lends. DUR PRICE £7.00

MODEL HIDKI 730X

30,000 opv. Over load protection 6 30 60 300 600 1200V DC. 12 60 120 600 1200V AC 120 600 1200 60 μΑ 30mA 300mA 2K 200K 2 Meg Ohm.

OUR PRICE £7.50 P&P 15p

MODEL TE300 MODEL TE300 30,000pp. Mirror saile. Overload protection. 0/0.6/3/15/ 0/300/1200V DC. 0/6/30/1200V DC. 0/6/30/1200/600/ 1200V AC. 0/30uA/ 6mA/60mA/30mA/ 600mA. 0/8k/80k/ 800k/8 Meg ohms. -20 to +63dB.



0.3/3/30/300mA/ 3A AC. Resistence: 25/500 ohms/0.5/5/50/500k ohms/5 Mohms. Decibels: –10 to +12dB. Size 167 x 98 x 63mm. Supplied comp-lete with test leads, spare diode and instructions.

OUR PRICE £8.00 P&P 20p U435 MULTIMETER

U435 MUL TIMETER
20,0000pv. Overload
protected, Ranges:
75mV/2.5/10/25/
100/250/500/1000V
C, 25/10/25/100/5/
C, 25/10/25/100/5/
Eurrent: 50uA/15/
25/100mA/0.5/2.5A
DC. 5/25/100mA/
0.5/2.5A AC, Resistance: 0.3/3/3/3/3/30/5

OUR PRICE £8.75 U91 Clamp VOLT AMMETER For measuring AC voltage and current without breaking circuit, Ranges: 300/600V AC. Current: 10/25/100/250/500A. Accuracy 4%, Size 283 x 94 x 36mm. Complete with carrying case, leads and fuses.

DUR PRICE £10.50 P&P 20p

P&P 17p

TMK MODEL TW50K

OUR PRICE £12.50

OUR PRICE £14.95

THE

OUR PRICE £17.50 P&P 40p

U4312 MULTIMETER

U4312 MULTIMETER
extremely sturdy
instrument for
general electrical
use, 657 ps. (57.50)
607.1557/300.6007
900V DC 8.75mV.
070.371.5/7.5/30
607.1557/300/6007
900V AC, 0/300uA/
1.5/67.157.150/60/
600mA/171.15/6A
DC, 071.56/150/60/
600mA/171.15/6A
DC, 071.56/150/

OUR PRICE £9.75 P&P 25p

MODEL 500 30,000 pp with overload protect-tion. Mirror scale, 0/0.5/2.5/10/25/ 100/250/500/ 1000/25/05/00/ 1000/25/05/06/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/100/ 250/06/ 250 MODEL 500

P&P 15p

UUR PRICE £ 13.95 Carr. paid Leather case for above £1.75

HIOKI 750X VOLT-OHM-MILLIAMETER

MILLIAMETER
43 ranges 0-0.3/0.6/
1.5/3/6/12/30/60/150/
300/600/1.2004
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DUR PRICE £11.95 P&P 40n

HIUKI MODEL 7000
100.000opv. Overtoad
protection. Micro scale.
12/30/60/120/300/
12/30/60/120/300/
15/3/6/12/30/60/150/
300/600/1200V AC.
15/30uA/3/6/30/60/
150/500mA/6/12A DC.
2k/200k/2M/20MOhms.
-20 to 463dB. HIOKI MODEL 700X +63dB

Model HT10084 MULTIMETER

MODEL AS. 1000 VOM

100,000 opv. Mirror scale. Mirror scale.
Built-in meter
protection. 0, 3
12 60 120 300
600 1200V DC.
0 6 30 120 300
600 VAC. 0 10µA
6 80 300mA
12 Amp. 0 2K
200K 2 M 200 Meg
Ohm. 20 to 17 dB

DUR PRICE £17.50 P&P 20p.

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TMK 100K LAB TESTER

TMK 100K LAB TESTER
100 000-pv. 6.7"
scale, Burer check,
Sensitivity 100,000
opy DC, 5k/V AC
DC Volts: 0.5/2.5/
0.750/250/1000V
AC, 3/10/50/250/1000V
AC, 3/10/50/250/1000V
AC, 3/10/50/250/100 Ac,
Southern 10/100/Meg/100 Meg ohms,
Decibels - 10 to +4986, Plastic case
with carrying handle. Size: 190 x 172
x 99mm.

OUR PRICE £19.95 P&P 25p 370WTR MULTIMETER

370WTR MULTIMET Features AC current ranges. 20.000pov 0/0.5/2.5/10/50/ 250/500/1000V DC. 0/0.5/2.5/10/250/ 500/1000V AC. 0/50.4/11/0.4 DC. 0/100mA/1/10A DC.

OUR PRICE £19.95 P&P 25p KAMODEN 72,200 Multitester

KAMD DEN 72, ZUV N High sensitivity tester. 200,000 ap V Overload protected Mirror scale. G6/.3 1200 V DC. 0/3 12/60/300/11200 V AC. 0/6uA/ 12/m2/120m/ 12/m2/120m/ 63dB. 0/26/200k/ 2 Meg/200 Megohms. UR PRIECE 522.5

OUR PRICE £22.50 P&P 30p

U4317 MULTIMETER

U4317 MULTIMETER
High sensitivity
instrument for field
and laboratory work.
Knife edge pointer.
Senm. mirror scale.
Ranges: 100mV/
0.5/2.5/10/2.5/5.0/100/25/5.0/100/0.
500/1000V. AC. Current 50. A.0.5/
15/10/50/250mA/17/5.A DC. 0.25/
0.5/10/50/250mA/17/5.A AC. Resistance: 0.5/10/100/200 ohms/13/
30/300k ohms. Decibels: -5 to 10dB
Battery operated. Size: 210 x 115 x

OUR PRICE £15,00 P&P 20p

MODEL U4311 Sub-standard

7.5/15/30/75/ 150/300/750mA/ 1.5/3/7.5A AC. 0/75/150/300/750wV/1.5/3/7.5/15/ 30/75/150/300/750V DC. 0/750mV/ 1.5/3 /7.5/15/30/75/150/300/750V AC. Automatic cut out device. Supplied complete with test leads, manual and test certificates.

OUR PRICE £49.00

TE65 VALVE VOLTMETER 28 ranges, DC volts 1.5-1500V, AC volts 1.5-1500V Resistance up to 1000 Megohms, 200/240V AC operation, Com-plete with probe and instructions. OUR PRICE £17.50 P&P 30p

Additional probes available: RF £2.12, HV £2.50

MODEL AF.105 VDM 50,000 opv Mirror scale Meter protection 0 3 3 12 60 120 300 600 1200V DC. 0 6 30 120 DC. 0 6 30 120 300 600 1200 V DC. 0 30μΔ β e i 60 300 mA 12 Amp. 0 10K 1m 10m 100 Meg Ohms 2 20 10 17 dB

P&P 20p.

OUR PRICE £12.50

LB3 TRANSISTOR TESTER

Tests ICO and B. PNP/NPN. Operates from 9V battery. Instructions supplied. OUR PRICE £3 95 P&P 20n

LB4 TRANSISTOR TESTER

TESTER
Tests PNP or NPN
transistors, Audio
indication, Operates
on two 1.5V
batteries, Complete
with instructions etc. **OUR PRICE** €4.50 P&P 20p



U4341 Multimeter & Transistor Tester

DUR PRICE £10.50

STOOTR MULTIMETER TRANSISTOR TESTER

TRANSISTOR TE 100,000opv, Mirror scale, Overload protection, 0/0,12/ 0.6/3/12/30/120/ 600V DC, 0/6/30/ 120/600V AC, 0/12/600uA/12/ 300mA/6/12A DC 0/10k/1 Mag/ -20 to +50dB, 0,01-0,2 MFD Transistor tester mea

O.01-0.2 MFD
Transistor tester measures Alpha, Beta and ICO. Complete with instructions, batteries and leads.

OUR PRICE £19.95 P&P 25n

KAMODEN HMG500 insulation resistance tester

INSUI at 100 Hesistanica teator Range 0 - 1,000 Megohms, 500V. Battery operated. Wide range clear meter 4" x 4". Complete with deluxe carrying case, batteries and instructions.

OUR PRICE £19.95

P&P 30n

CIS PULSE OSCILLOSCOPE

CIS PULSE OSCILLOSCOPE
For display of pulsed
and periodic wave
forms in electronic
circuits, VERT, AMP.
Bandwidth: 10MHz,
Sensitivity at 100kHz
VRMS/mm: 0.1–25;
HOR. AMP. Bandwidth: 500kHz
VRMS/mm: 0.7–80e
For pulse of the p

OUR PRICE £39.00

RUSSIAN CI16 Double Beam

OSCILLOSCOPE OSCILLOSCOPE

5 MHz pass band.
Separate Y1 and Y2

mplifiers. Rectangular 5'' x 4'' CRT.
Calibrated triggered
sweep from 0.2use.
to 100 milli-sec/cm.
Free running time
base, 50Hz-1MHz.
Built-in time base
Calibrator and amplitude Calibrator.
Supplied complete and amplitude Calibrator.

DUB PRICE £87.00 Carr, paid

MODEL TE15

MUUEL 1E15
GRID OIP METER
Transistorised. Operates as Grid Dip,
Oscillator, Absorbtion Wave Meter and
Oscillating Oetector. Frequency range 440kHz = 280MHz in six coils, 500uA meter, 9V battery operation, Size: 180 x 80 x 40mm. OUR PRICE £19.95



Also see following pages

ALL PRICES EXCLUDE VAT

P&P 20a

SWR METER Model SWR3

SWM METER Model S Handy SWR meter for transmitter antenna align-ment, with built-in field strength meter. Accuracy 5%, Impedence 52' Indic-ator 100uA DC. Full scale 5 section collapsible antenna. Size 145 x 50 x 60mm.

AT201 Decade

ATTENUATOR

OUR PRICE £12.50

TRANSISTORISED L.C.R. A.C.

A new portable bridge offering

bridge offering
excellent range and
accuracy af low
cost. Resistance.
6 ranges: 1 microhenry-111
henries ± 2% Capacity. 6 ranges
10pf-1110 mfd ± 2 Turns Ratio:
6 ranges: 1.11/1000-1.11100 = 1%
Bridge Voltage at 1.000cr Opers

6 ranges: 1:1/1000-1 11100 = 1% Bridge Voltage at 1,000cps. Opera-ted from 9-volt battery. 100 micro-amp meter indication. Size 78" x 5" x 2" OUR PRICE £25.00 P&P 25p

TE16A TRANSISTORISED

TE16A TRANSISTORIS
SIGNAL GENERATDR
5 ranges, 400kHz
to 30 MHz. An
imaxpensive for
the handy-man.
Operates on 9V
battery. Wide
easy to read
scale. 800kHz
modulation.
Size: 149 x 149 x 92mm.
with instructions and leads.

OUR PRICE £8.97

GENERATOR

OUR PRICE £17.50

TE-200 RF SIGNAL

Accurate wide range signal generator covering 120 kHz-500 MHz on 6 bands. Directly callbrated. Variable R.F.

for calibration, 220/240V

attenuator audio output. Xtal socker

OUR PRICE £17.50 P&P 30p

TE22 SINE SOUARE WAVE AUDIO GENERATOR

AC operation, Supplied brand new guaranteed, with instruction manual and leads

OUR PRICE £24.95 P&P 37p

.0

P&P 50o

ARF 300 AF/RF SIGNAL

GENERATOR

All transistorised

All transistorised compact fully portable. AF sine-wave 18Hz to 220 kHz. AF square wave 18Hz to 100k Hz. Output Square/ Sine wave 10V. P.P. RF 100kHz to 200MHz. Output 1V maximum

OUR PRICE £37.50

GENERATOR

Sine 20cps to 200kHz

on 4 bands Square 20 Square 20 cps to 30 kHz. Output

impedeno 5000 Ohri 200/250V AC opera

MODEL TE20 RF SIGNAL

GENERALUM Six bands. 120kHz— 260MHz. Dual output RF terminals. Separate variable audio output. Accuracy ± 2%. Audio output to 8V. Power requirements. 105—125V. 220-240V AC. Size: 193 x 265 x 150mm. Complete with test

92mm. Complete d leads.

P&P 40m

.

BR/8 MEASURING BRIDGE

OUR PRICE £4.25 P&P 250



MODEL MG 100 SINE SQUARE WAVE AUDIO GENERATOR Range 19-220,000Hz Sine

Wave 19—100,000 Hz Square Wave Output Sine or Square wave 10v P to P Size 180 x 90 x 90mm Operation 220/240v. A.C. OUR PRICE £19.95 PR P 3 7m

SUPPLY UNIT Solid state. Variable output 5-20V DC up to 2 Amp. Independent meters to monitor voltage and current. Output 220/240V AC. Size: 190 x 136 x 98mm. Frequency range 0— 200kHz. Attenuator 0—111dB, 0.1dB steps. Impedence 600 ohms. Input power maximum 30dBm. Size: 180 x 90 x 55mm.

P&P 37n



PS200 Regulated POWER

OUR PRICE £19 95

POWER RHEDSTATS High quality ceramic construction. Wind ings embedded in ings embedded in Heavy duty brush wiper. Continuous rating. Wide ramper available ex-stock. Single hole fixing. Wide diameter shafts. Bulk quantities available.

25 WATT 10/50/100/250/500/ £1.15 P&P 10p 1000 Ohms 58 WATT 10/25/50/250/500/ 1000/2500/5000 Ohms. £1.62 P&P 10p

100 WATT 1/10/25/50/100/250/ 500/1000/2500 Ohms

£2.34 P&P 15p

YAMABISHI VARIABI E **VOLTAGE TRANSFORMERS**

VOLTAGE TRANSFORMERS
Excellent quality at 100w cost. Input:
230V 50/60Hz. Output 0 – 250V
MODEL S260 BENCH MOUNTING
PSP
1A £10.50 30p
2.5A £12.00 35p
5A £17.50 37p
8A £30.35 50p
10A £33.75 75p
12A £29.50 75p
20A £85.00 125p
20A £95.00 130p
MODEL S260B PANEL MOUNTING
1A £10.00 30p

1A £10.00 30p 2.5A £12.00 35p

BVD5 Vernier TUNING DIAL

App. 7:1 ratio planetary drive vernier dial. Log scale 0—180 degrees. Blank scales 1—5. Dial size 128 x 76mm. Overall size 190 x 117 x 14mm. deep including knob and coupling. ³⁶ diam. shaft OHR PRICE £1.62 P&P 15r

240° Wide Angle 1mA METERS MW 1-6 60x60mm

£6.50 P&P 15p MW 1-8 80x80mm £6.90 P&P 15p



CP110 CHASSIS PUNCH SET

9900 P

Carefully machined top grade steel. Contains 1/2", 5/8", 3/4", 1" and 1 1/8" punches complete with gripper and accessories. **OUR PRICE £3.00** P&P 40p

HITACHI FLUDRESCENT LANTERN LIGOI A portable battery operated lantern ideal for home, motoring, camping etc. Approx. 10 tall. Provides brilliant light from 9 1.5v batteries (not

0 supplied) OUR PRICE £7.19 P&P 25p

KE630 3 Station INTERCOM



Master and two sub-stations. Can be used on desk or wall mounted. Complete with cable and batteries OUR PRICE 65 25 P&P 50n



LIMITED STOCKS OUR PRICE £24.95 per pair P302 Two Channel 300mW OUR PRICE £52.50 per pair 1003 Three Channel 1 Watt DUB PRICE £71.25 per pair

DT55G DIGITAL CLOCK DT55G DIGITAL CLOCK
MECHANISM
Features
24 hour
alarm
setting,
and auto
alarm 'Sleep' switch. Illuminated rotalarm 'Sleep' switch. Illuminated rotalarm 'Sleep' switch. Illuminated rotconds. Automatically turns off radio,
TV. light etc. and with auto-switching will turn on again when required.
240V AC operation. Switch rating
250V-3 Amp.

SKYFON 100mW

&P 50p per pair N.B. Unlicenceable in U.K.

DUR PRICE 65 95 P&P 30p

SINCLAIR IC12 INTEGRATED CIRCUIT AMPLIFIER ALVE complete with printed circuit

P&P 30p

P&P 30p

OUR PRICE £2.35 P & P 15p LH02S STEREO HEADPHONES

Light weight head-phones with padded ear pieces. 4/16 ohms 20-20,000Hz. De with 6' lead and plug OUR PRICE £1.97

TE1018 Deluxe Mona High

Impedence Headset. Impedence neause. Sensitive magnetic headset with soft ear pads. Impedence 2,600 ohms (600 ohms DC). Frequency response: 200–4,000Hz.

OUR PRICE £2,25 P&P 30s TE1035 Stereo HEADPHONES 100

LOW cost with excellent response. Foam rubber earcups. Adjustable headband. B ohms impedence. Frequency response 25Hz—18kHz. Complete with cable and stereo jack plug. OUR PRICE £2.60

SH8DV MONO/STEREO HEADPHDNES Volume control for each channel. 4/16 ohms impedence. Frequency response 20Hz – 18kHz. Complete with 10ft. coiled lead and jack plug. 106 plug. OUR PRICE 64 97

BH001 HEADSET and Boom Microphone
Moving coil, Ideal
for language
teaching,
community
cations etc.
Headphone impedence 16 ohms. Microphone impedence 200 ohms. Micronhone OUR PRICE £5.95 P&P 30p

EMILDUDSPEAKERS Model 350 13 x 8" with single tweeter/crossover. 20-20,000Hz. 15 watts RMS. Available 8 or 15 ohms. 00. OUR PRICE £7.50 each P&P 37p Model 450 13 x 8" with twin tweeter/crossover. 55—13,000Hz. 8 watts RMS. Available 8 or 15 ohms

OUR PRICE £3.62 each P&P 25p

SPECIAL PURCHASE LIMITED DUANTITY Tannoy 12" DR/8 Bass Speakers 8 ohms. 30 watt Heavy duty, ideal for Hi-Fi P.A.

Group. OUR PRICE £12.50.

SPECIAL BARGAIN! FERGUSON 3406 HI-FI SPEAKERS

High quality 2 way speaker systems. 25 Watts, 4-8 ohms, 40Hz-18kHz. Size: 560 x 340 x 255mm, approx. Wood grain finish with black fronts. OUR PRICE £26.95 PR. P&P £1

SPECIAL BARGAIN !! STEREOSOUND SPEAKERS

Matched pair of stereo bookshelf speakers. Deluxe teak veneered finish. Size: 368 x 229 x 190mm. 8 ohms. 8 watts RMS, 16 watts peak. Complete with Din lead.



OUR PRICE £12.95 FM TUNER CHASSIS

6 transistor high quality high quality tuner. Size only 153 x 101 x 63mm 3 IF stages. Double tuned discriminator.

discriminator. Ample output to feed most amplifiers. Operates on 9V battery. Covers 88– 108MHz. Ready built, ready for use.

OUR PRICE £8.95 Stereo Multiplex Adaptor 65 95 p. 169

Model A1018 **FM TUNER** 6 transistor high quality unit— 3 IF stages and double tuned

double tuned discriminator. For use with most amplifiers. Covers 88-108MHz. Powered by 9V battery. OUR PRICE £13.50 P&P 30p Stereo multiplex adapter £5.95 extra

SINCLAIR "SCIENTIFIC" CALCULATOR 8 digit display. Four functions plus logarithms to base 10, antilog, sine, cosine, tangent, arcSine, arcCosine and arcCosme and arcTangent. Complete with instructions, case and batteries. Rec. Price

OUR PRICE £44.50 P & P 25p plus VAT

SINCLAIR SYSTEM 2000 STEREO AMPLIFIER AND TUNER

70000 TOOO **AMPLIFIER**

Amplifier output 8 watts per channel RMS. Distortion less than up % Silicon transistors. Two ick-up plus radio and tape inputs apperoutput and scratch filter xcellent Value Silicon transistors.

OUR PRICE £28.50 P&P60p

■ 🖨 33

EM TIINER

Excellent selectivity and sensitivity. Twin dual-varicap tuning. 4 pole ceramic filter. 19 transistor stereo demodulator giving 40 dB separation. Distortion 0.2% output. Fantastic Value.

OUR PRICE £28.50 P&P60p.

SINCLAIR Project 80 Modules 240 Power Amplifier.
Z60 Power Amplifier.
Z60 Power Amplifier.
Stereo 80 Pre-Amplifier.
Active Filter Unit.
Project 805.
PZ5 Power Supply.
PZ6 Power Supply.
PZ8 Power Supply.
Transformer for PZ8. £5.45 £6.95 £11.95 £6.95 £26.95 £4.98 £7.98 £7.98 £4.05 SINCLAIR Project 80 Packages 2 x Z40/Stereo 80/PZ5. 2 x Z40/Stereo 80/PZ6.... 2 x Z60/Stereo 80/PZ8... POST & PACKING 35p each

MP7 MIXER-PREAMPLIFIER MP7 MIXENTREON

Shirophone
Inputs each with
Individual gain
controls enabling
complete mixing
facilities. Battery operated. Size: 235
x127 x 76mm. Inputs: Mics. 3 x 3mV
50K; 2x 3mV 600 ohms. Phono. Mag.
4mV 50K; Phono Ceramic 100mV 1
Meg. Output 250mV 100K.

Par Parier 68 q17
P&P 20p.

DUR PRICE £8.97 P&P 20p



HIGH QUALITY CONSTRUCTION KITS WE ARE APPOINTED STOCKISTS AT

ALL BRANCHES All kits are complete with comprehensive easy to follow instructions and covered by full guarantee.

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Post and Packing 15p per kit.

AF20 Mono amplifier. C3.60

AF28 Miser grampilifier. C3.60

AF28 Miser grampilifier. C3.60

AF28 Miser grampilifier. C3.60

AF28 Miser grampilifier. C5.91

AF3 Mono amplifier. C5.92

AF3 Mono amplifier. C5.92

AF3 Mono amplifier. C5.93

AF3 Mono a

Amateur Electronics by Jossy-Kit, the professional book for the amateur covers the subject from basic principals to advanced electronic techniques. Complete with circuit board for AE1 to AE10 listed below.

OUR PRICE £3.30 (No VAT) P&P 25p plus VAT

P&P ZSp plus VAT.

AE1 100mW output stage.
AE2 Pre-amplifier
AE3 Diode receiver.
AE4 Flasher.
AE5 Astable multi-vibrator.
AE6 Monostable multi-vibrator.
AE7 RC generator.
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50u	Α			£380			
100				£3 75		_	74
200				£3 70			
500				£3 65	3.00		
	0-50u			£3 75		Α	
100	-0-10	Ou A		£370		-	- 0
1m	Α			£3 65			393
5m	Α			£3 65			
	nΑ			£3.65	Carlot and Carlot		0.040
50n	nΑ.,			£3.65	10V DC		
100	lmΑ			£3.65	20V DC		
500	mΑ		.,	£3.65	50V DC		
1A	DC			£3 65	300V DC		
5A	DC			£3 65	15V AC		
104	A DC	٠.		£3 65	300V AC		
5V	DC			£3 65	VU Meter		

CLEAR PLASTIC MODEL SW100 Size: 100 x 80mm

50u A	1460	F.	
100uA	£4.50		- 1
500u A	£4 30	Α	- 11
50-0-50u A	£4.50	-	- 11
100-0-100u A	£4.45	Annual Property of	-
1mA	£4 30		
1A DC	£4 30	The state of the s	
5A DC	£4 30		
20V DC	£4 30	150V AC	£4 45
50V DC	£4 30	300V AC	£4 45
300V DC	£4 30	VU Meter	£4 90

EDGWISE MODEL PE70

50uA	 £4 15
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200u A	£4 05
500u A	£3 90
50-D-50u A	 £4 10
100-0-100u A	 £4 05
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100u A	£7 90		
50-0-50u A	£7 90	CDL/ DC	-3.50
l m A	€7.60	2DV DC	£760
		50V DC	£760
1-0-1mA	£7 60	300V DC	£7 60
1A DC	£760	500mA/5A DC	£8 60
5A DC	£760	5V/50V DC	€8 60
5V DC	£7.60	5V/15V DC	£B 60
10V DC	£760	1/5A DC	£8 60
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CLEAR PLASTIC MODEL MR 85P

50uA		 £5	45	
100u A		 £5	40	
200u A		 £5	35	
500u A		 £5	25	
50-0-50u A	A .	 £5	40	1 6
100-0-100		£5	35	1 .
500-0-500	λuA	 £5	20	26.1
1m A		 £5	20	limum
1-0-1mA		 £5	20+	¥
5mA		 £5	20	
10mA		 £5	20	
50mA		 £5	20	
100m A		£5	20	300V
500mA		£5	20	15 V A
1A DC		 £5	20	300V
5A DC		£5	20	S Mete
15A DC		 £5	20	VU Me
30A DC		 £5	40	1A AC
10V DC		65	2.0	5A AC



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)			_		
)	300V DC			£5	
)	15V AC			£5	3
)	300V AC			£5	3
)	S Meter 1mA			£5	2
)	VU Meter			£5	5
)	1A AC .		٠	£5	2
)	5A AC		•	15	2
)	10A AC		٠	15	2
)	20A A C		•	15	
'n	30A AC		*	15	2

*Items with asterisk are Moving Iron type, all others are Moving Coil

CLEAD DI ACTIC MODEI CD020

ULEAN PL		WUL	LE PDO	טנ		
Size: 110 x	83mm					
50u A		30	-		 	_
100u A		25				1
200u A		20				Ш
500u A		15		Α		Ш
50-0-50u A		25		-		Ш
- 100-0-100u /		20		-		Ш
1mA		10				8
5mA		10		ĸ.		8
10mA		10	2000000			_
50mA		10	10V DC		 £4	
100mA	£4		20V DC		 £4	
500mA			50V DC		 £4	
1A DC	£4	10	300V DC		 £4	
5A DC		10	15V AC		 £4	
10A DC			30DV AC		 £4	
5V DC	. £4	10	VU Meter		£4	40

CLEAR PLASTIC MODEL MR 45P

Size: 50	x 50)mm	
50u A			£3 20
100u A			£3 15
200u A			£3 10
500u A			£3 00
50-0-50u	Α		£3 15
100 0 10	Ou A	۸	£3 10
500-0-50	Ou A	۸	£2 95
1mA .			£2 95
5m A .			£2 95
10mA			£2 95
50mA			£2 95
100m A			£2 95
500m A			£2 95
1A DC			£2 95
5A DC			£2 95
10V DC			£2 95
20V DC			12 95



CLEAR PLASTIC MODEL MR 38P

1	50u A .		£3 10
i	100u A		£3 05
	200u A 500u A		 €3 00
	500u A		 £2 85
	50 0-50u A	4	 £3 05
	100-0-100	u A	 £3 00
	500-D-500	λuΑ	£2 80
Į	1mA		£2 80
ł	1 0-1 mA		£2 80
	2mA		£2 80
	5mA 10mA		£2 80
Į	10mA		£2 80
	20mA		£2 80
ł	50mA .		£2 80
	100mA		£2 80
ł	150mA		£2 80
ł	200mA		£280
Į	300mA		£2 80
	500mA		£2 80
	750mA		£ 280
	1A DC		£280
	2A DC		£2 80
	5A DC		£2 80
	10A DC		£2 80
l	3V DC		£2 80
	10V DC		£2 80
	15\/ DC		62.90



CLEAR PLASTIC MODEL SD460

Size. 59 x 46mm			
50u A	£3 50		
100u A .	£3 45		
200u A	£3 40	i i	
500u A .	£3 35	1.3	- 11
50-0-50u A	£3 45		Δ
100-0 100uA	£3 40		
1mA	£3 30	-	The same of
5m A	£3 30		
10mA	£3 30		
50mA .	£3 30	10V DC	£3.30
100mA	£3 30	20V DC	£3 30
500mA	£3 30	50V DC	. £3 30
1A DC	£3 30	300V DC	£3 30
5A DC	£3 30	15V AC .	£3.45
10A DC	£3 30	300V AC	£3 45
5V DC	£3 30	VII Meter	£3 65
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BAKELITE MODEL S80 Enlarged Window

50u A	£4 50	
100u A	£4 45	-
500u A	£4 20	
50-0-50u A	£4 45	1
100-0-100uA .	£4 40	The second second
1mA	£4 20	
1A DC	£4 20	1 / "
5A DC	£4 20	-
20V DC	£4 20	10.00
50V DC	£4 20	
300V DC	£4 20	
300V AC	£4 30	
VU Meter	£4 70	

CLEAR PLASTIC MODEL MR 52P

126. OO x OO		
00uA 00uA 00uA 00-0 50uA 00-0 100uA ImA 0mA	 £3 70 £3 50 £3 35 £3 50 £3 45 £3 30 £3 30 £3 30	MA
00mA .	£330	
00mA .	£3 30	WITH THE PARTY OF
ADC .	£3 30	
A DC .	£3 30	S.Meter 1mA
OV DC	 £3 30	VU Meter
OV DC .	£3.30	1A AC *



25uA	£5 25	CONTRACTOR OF STATE
50uA	£4 00	
100u A	£3 95	
500u A	£3 65	
50 0-50u A	£3 95	10 hA
100-0-100u A	£3 90	
500-0-500u A	£3 60	THE PARTY NAMED IN COLUMN TWO IS NOT
1mA	£3 60	
1-0-1mA	£3 60	
5m A	£3 60	
10mA	£3 60	
50mA	£3 60	300V DC £3 60
100mA	£3 60	30V AC * £3 60
500m A	£3 60	50V AC * £3 60
1A DC	£3 60	150V AC * £3 60
2A DC	£3 60	300V AC * £3 60
5A DC	£3 60	500V AC £3 60
10A DC	£3 60	VU Meter £4 10
15A DC	£3 60	1A AC * £3 60
30A DC	1360	5A AC 1360
50A DC	£3 80	10A AC 1360
5V DC	£3 60	20A AC 13 60
10V DC	£3 60	30A AC . * £3 60
15V DC	f3 60	50A AC
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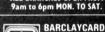
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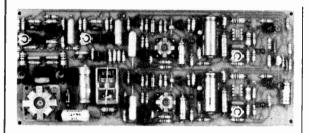
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(PE Nov 70/Mar 71) Stereo Sets and PCBs Pre-amp—Rs. Ca. Pots. Sw s—with JW MO Rs £14-18—with JW CF Rs. £10-40. PCB as published, £2-20. Main Amp—Rs. Cs. Pots. £5-88. PCB (3½n × 5in). £1-28. Power supply—Rs. Cs. Pot. £4-56. PCB (2in × 4in).

SOUND SYNTHESISER

(PE Feb. 73/Feb. 74)

RHYTHM GENERATOR (PE Mar./June 74)

SOUND BENDER

(PE May 74)

Details of all these in List

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(PW Nov./Dec. 72) S/c s. Rs. Cs. T/former
—with Rotery Pots. £6-44 PCB (2in × 11½in), £1-40, 9in Spring Unit, £4-50.

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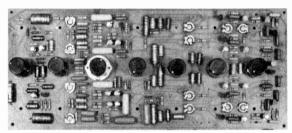
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PROJECT 04

(PW Oct 73 Jan 74) Multisystem Quadraphonic Decoder. Sic's, I/c's, Rs. Cs. Pots. Makeswitches, £12 48 PSU, £3-17. Set of PCBs, £2-60.

PHASING UNIT

(PE Sept 73) S/c's, Rs, Cs, Pots, PCB (1\frac{1}{2}\in \times 2\frac{1}{2}\in), £2-20.



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(PE Apr./Aug 71) Multichannel Sound Controlled Light S/c s (Excl SCRs), Rs. Cs. Pots.
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ELECTROLYTIC (µF/V)

100 40 100 63 150 16 150 63 220 10 220, 16
220, 25
220, 16
220, 25
220, 40
220, 63
330, 10
470, 6, 3
470, 10
470, 25
470, 40
500, 64
880, 63
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1000, 10
1000, 16
1000, 16
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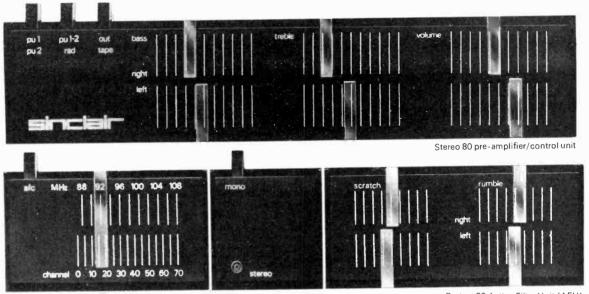
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Sinclair Project 80 exciting



Project 80 tuner

Stereo decoder

Project 80 Active Filter Unit (AFU)

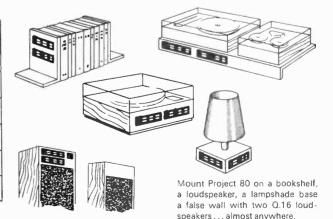
only $\frac{3}{4}$ " deep x 2" high

Living with hi-fi takes on new meaning with Sinclair Project 80. The electronics of these revolutionary new modules are all contained within elegantly designed matching cases no more than three-quarters of an inch deep. They are designed for mounting on any appropriate flat surface by means of 6BA bolts extending from the rear of each module and which pass through suitably drilled holes. Connections are taken away out of sight in a similar manner. The possibilities opened up by Project 80 are endless – superb hi-fi systems can be installed in ways hitherto only dreamed about and never before made practical. No more cutting out and shaping to put modules in position. A few holes drilled with the aid of templates supplied and the job is done. Now you need never again be faced with problems of keeping the hi-fi from clashing with carefully thought-out furnishing schemes. (That will surely please wives) Slider controls have been introduced in place of knobs and all modules in the range incorporate new up-dated circuitry with emphasis on performance standards and built-in protection against overload and shorting. The aim was to re-think modular construction completely – to make it infinitely more versatile, even simpler and more reliable – the result – Project 80 – another triumph for Sinclair, and the most exciting construction modules ever.

the slimmest, most elegant hi-fi modules ever made

System	The Units to use	Units cost
Simple battery record player	Z.40	£5-45 +54p V.A.T
Mains powered record player	Z.40, PZ.5	£10-43 +£1.04 V.A
30W. RMS continuous sine wave stereo amp.	2 × Z.40s, Stereo 80; PZ.6	£30 83 +£3.08 V A
50W (8 Ω) RMS continuous sine wave de luxe stereo amp.		£33·83 +£3.38 V.A

Project 80 FM tuner, decoder, and A.F.U. may be added as required



new thinking in modular hi-fi

Stereo 80 pre-amplifier and control unit

For P.U., radio and tape

 Tape monitoring switch Simplest ever fixing

Each channel has its own separate tone and volume controls operated by sliders, enabling ideal environmental matching to be obtained. A virtual earth input stage forms part of the up-dated circuitry that ensures the finest possible quality from all signal sources. Generous overload margins are allowed on all inputs. Clear instructions with template are supplied.

TECHNICAL SPECIFICATIONS

Size $-260 \times 50 \times 20$ mm $(10\frac{1}{4} \times 2 \times \frac{3}{4}$ Ins) Finish – Black with white indicators and transparent Sliders

Inputs - Magnetic pick-up 3mV RIAA corrected; Ceramic pick-up 300mV

Racio 300mV; Tape 30mV Signal/noise ratio – 60db

Frequency range - 20Hz to 15KHz ± 1dB. 10Hz to 25KHz + 3dB

Power requirements – 20 to 35 volts
Outputs – 100mV + AB monitoring for tape

Controls - Press button for tape radio and P.U. Sliders for volume bass (\pm 12dB to - 14dB at 100Hz) treble (\pm 11dB to - 12dB at 10KHz)

R.R.P. £11.95 $^{+£1\cdot19}_{V.A.T}$

Project 80 FM tuner

and stereo decoder





■ Twin dual varicap tuning: ■ On the decoder, solid 4 pole ceramic filter: switchable A.F.C.

state stereo indicating beacon.

Making the Project 80 F.M. tuner and decoder available separately gives a wider choice of systems and saves money where stereo reception may not be required. The tuner is a triumph of electronic design and assures excellent performance. The decoder gives a 40dB channel separation with 150mV output per channel. Both units may be used with other than Project 80 systems

TECHNICAL SPECIFICATIONS OF TUNER

Size $=85 \times 50 \times 20$ mm $(3\frac{1}{2} \times 2 \times \frac{3}{2} \text{ins})$ Tuning range =87.5 to 108 MHz Detector =1.C balanced coincidence for good A.M. rejection

One I C. equal to 26 transistors
Distortion – 0.2% at 1 KHz for 30% modulation

4 pole ceramic filter in I.F. section

Aerial impedance – 75 Ω or 240-300 Ω Sensitivity – 4 microvolts for 30dB queting

Output - 300 mV for 30% modulation

Power requirements - 23 to 33 volts DECODER

Size $-47 \times 50 \times 20$ mm $(1\frac{7}{8} \times 2 \times \frac{3}{4}$ ins)

One 19 transistor I.C.

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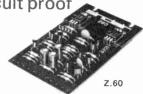


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Practical Electronics July 1974

Z.40 & Z.60 power amplifiers totally short-circuit proof





Intended for use in Project 80 installations, these modules readily adapt to an even wider range of applications. Both incorporate built-in protection against short circuiting and risk of damage from mis-use is greatly reduced

Z.40 TECHNICAL SPECIFICATIONS

Size $-55 \times 80 \times 20$ mm ($2\frac{1}{8} \times 3\frac{3}{8} \times \frac{3}{4}$ Ins) 9 transistors Input sensitivity -100mV

Output – 15 watts RMS continuous into 8 Ω (35v)

Frequency response - 10Hz - 100KHz ± 1dB Signal/noise ratio ~ 64dB

Distortion – at 10 watts into 8 Ω less than 0.1% Power requirements - 12 to 35 volts

Z.60 TECHNICAL SPECIFICATIONS

Size $-55 \times 98 \times 15$ mm ($2\frac{1}{8} \times 3\frac{3}{4} \times \frac{3}{4}$ Ins) 12 transistors Input sensitivity $-100 \cdot 250$ mV

Output – 25 watts RMS continuous into 8 Ω (45V). Distortion - typically 0.03%

Frequency response - 10Hz to more than 200KHz ± 1dB

Signal/noise ratio - hetter than 70dB

Built-in protection against transient overload and short circuiting

Load impedance -4Ω min, max-safe on open circuit Z.40 R R.P. £5.45 + 0.54 V A T , Z.60 R R P £6.95 + 0.69p V.A.T

Project 80 active filter unit

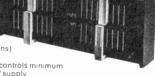
Makes a highly desirable part of any worthwhile system where inputs may be from record, radio or tape. As with Stereo 80, separate controls applied to each channel make it easier to obtain ideal stereo balance.

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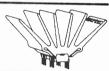
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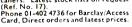
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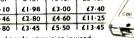
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