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Our April issue will be published on Friday, March 8, 1974

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#### Abstract



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ADI61 AD162 AFI 17 AF1 18
AFI 39 AF139
AF186 AF239
A5Y27 A5Y2
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AMP


H $\mathrm{H} \times \frac{1}{2} \mathrm{~L} \times$ hin．dia．
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2 AMPS
82
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82
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| B4／100 | 100 V | 55p |  |
| :---: | :---: | :---: | :---: |
| B4／200 | 200 V | 59p |  |
| B4／400 | 400 V | $65 p$ |  |
| B4／600 | 600 V | 75 p | ／ |
| B4／800 | 800 V | 41.00 |  |
| $\frac{1}{4} \mathrm{H} \times \frac{1}{2}$ | din．dia． |  | 11 |


| 6 AMPS |  |  |
| :--- | ---: | ---: |
| B6／05 | 50 V | 65 p |
| B6／10 | 100 V | 70 p |
| B6／20 | 200 V | 80 p |
| B6／40 | 400 V | 90 p |
| B6／60 | 600 V | 1.00 |

$\mathrm{t} H \times \frac{1}{2} \mathrm{~L} \times$ Ain．dia．
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Trpe
IAMP（TO5）P．I．V．I
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CRS I／1OAF
CRS I／20AF
CRS I／40AF 400 V
3 AMP（TO48）
CRS 3／05AF
CRS 3／10AF
CRS 3／20AF
CRS 3／40AF
5 AMP
7 AMP（TO48）
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$\begin{array}{lll}\text { CRS } 7 / 600 & 600 \mathrm{~V} & 95 \mathrm{p}\end{array}$

| Mk III Sound to Light Unit Chassis Version <br> The audio drive voltage is derived directly from the amplifier output or across the speaker load The unit converts the audio-frequency signals into a three-coloured light display when used with coloured flood lamps or similar; the colour depending on the frequency of the signal and the intensity on the loudness of the audio source. <br> Max. power 1.5 kW per channel at 240 V a.c. <br> Complere assembly built and tested. Size: 9 in $\times$ <br> 7in $\times$ 3ín. Price $\mathbf{4 2 7 . 5 0 .}$ | MN3 3 Channel <br> I.C. Mixer <br> Kit <br> Three channel I.C. audio mixer kit MN3 is specially designed for the home coristructor or educational project. Suitable for use in domestic audio, discotheque and simple p.a. applications. The MN3 mixer kit has the following features: Advanced design using 5 ICs; Stider fader level controls mounted directly to p.c. board; Full range Bass and Treble contrals; Uses top grade components with fibreglass ready-drilled and tinned printed circuit board and the unit may be operated from twin 9 V batteries (PP3) if required. <br> As optional extras, an attractive ready-punched facia panel and control knobs are available. The unit is available as a ready built mixer with facia. <br> Size: $9.5 \mathrm{in} \times 4$-8in $\times 1.5 \mathrm{in}$ (with facia). <br> Construction manual available separately at 25 p. Kit Price Cl 1. | STL/I Single Channel Sound to Light Unit <br> Single channel "Sound to Light" unit with 60 mm slider fader controls for audio trigger level and background load dimmer. <br> This unit is switchable for response to High, Mid. and Low frequency audio signals, selected by facia control. A D.1. "pulse-flash" push button is fitted for manual flashing of the lamp load together with neon load indicator. <br> Maximum Load: IkW at 240 V a.c. <br> Size: $\operatorname{Bin} \times 3 \cdot 6$ in $\times 3 \mathrm{in}$. <br> Price $\mathbf{4} 14.30$. |
| :---: | :---: | :---: |
| Practical Electronics "Scorpio" Electronic Ignition System Kit <br> This Capacitor-Discharge Electronic Ignition system was described in the November and December issues of Pracsical Electroniss. it is sursable for incorporating in any l2V ignition system in cars, boats, go-karts, etc. <br> Case size: $7 \cdot 25$ in y $4 \cdot \operatorname{Sin}$, 2in approx. <br> Complere assembly and wiring manual 25p. <br> Price \$12.10. | STL/3 Sound to Light Unit <br> The performance of this unit is similar to the Mk III chassis version above, but is provided with an attractive black anodised faciz panel and designed to mount directiy into equipment desks or cabinets. High, Mid. and Low frequency channel sensitivity controls employ 60 mm slider faders for presentday attractive appearance and ease of use. Size: <br> 11 in $\times 8 \cdot 5$ in $\times 4$ in. <br> Price $\mathbf{E 4 2 . 3 5}$. | AUDI <br> EFFECTS <br> CREATE "PHASE"EF FECT ON YOUR RECORDS. TAPES, ETC., UNIQUE CIRCUITRY ENABLES YOU TO CREATE PHASE EFFECT AT THE TURN OF A KNOB. OPERATES FROM 9 V BATTERY (not supplied). COMPLETE KIT OF COMPONENTS WITH PRINTED CIRCUIT BOARD AND FULL INSTRUCTIONS. KIT PRICE 62.86. <br> All prices quored include postage and VAT. S.A.E. for further details. <br> DABAR ELECTRONIC PRODUCTS 98 LICHFIELD.STREET, WALSALL STAFFS WSI IUZ |



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| 307 | DIL |
| 307 | TO99 |
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| $320 \mu \mathrm{~F}$ | 818 | $150 \mu \mathrm{~F}$ | 8 p | $220 \mu \mathrm{~F}$ | 11 p |
| $1000 \mu \mathrm{~F}$ | 13p | $220 \mu \mathrm{~F}$ | 9 p | $470 \mu \mathrm{~F}$ | 10p |
| $4700 \mu \mathrm{~F}$ | 29p | $680 \mu \mathrm{~F}$ | 17 p | $680 \mu \mathrm{~F}$ | 26p |
| 6.3 VOLT |  | $1000 \mu \mathrm{~F}$ | 175 | $1000 \mu \mathrm{~F}$ | 25p |
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| $33 \mu \mathrm{~F}$ | $81 p$ | $2000 \mu \mathrm{~F}$ | 43p |  |  |
| $68 \mu \mathrm{~F}$ | ${ }^{610}$ | 25 VOLT |  |  |  |
| $150 \mu \mathrm{~F}$ | 61 D |  |  |  |  |
| $470 \mu \mathrm{~F}$ | 11p | $10 \mu \mathrm{~F}$ | $61 p$ |  |  |
| $680 \mu \mathrm{~F}$ | 18 D | $22 \mu \mathrm{~F}$ | ${ }^{615}$ | 63 VOLT |  |
| $1500 \mu \mathrm{~F}$ | 18p | $47 \mu \mathrm{~F}$ | 618 |  |  |
| $2200 \mu \mathrm{~F}$ | 18 p | $100 \mu \mathrm{~F}$ | 8 8 | $1 \mu \mathrm{~F}$ | 61p |
| $3300 \mu \mathrm{~F}$ | 26p | $150 \mu \mathrm{~F}$ | 8p | $2-2 \mu \mathbf{F}$ | 818 |
| 10 VOLT |  | $220 \mu \mathrm{~F}$ | 10 p | $4.7 \mu \mathrm{~F}$ | ${ }^{6 \frac{1}{2} \text { D }}$ |
|  |  | $470 \mu \mathrm{~F}$ | 13p | $6.8 \mu \mathrm{~F}$ | Bip |
| $22 \mu \mathrm{~F}$ | 61. | $680 \mu \mathrm{~F}$ | 20 p | $10 \mu \mathrm{~F}$ | 817 |
| $47 \mu \mathrm{~F}$ | 81 p | $1000 \mu \mathrm{~F}$ | 22p | $22 \mu \mathrm{~F}$ | 610 |
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| $220 \mu \mathrm{~F}$ | 8 p | $5000 \mu \mathrm{~F}$ | 68p | $100 \mu \mathrm{~F}$ | 11p |
| $330 \mu \mathrm{~F}$ | 10p |  |  | $150 \mu \mathrm{~F}$ | 18 p |
| $470 \mu \mathrm{~F}$ | 10 D | 40 V |  | $220 \mu \mathrm{~F}$ | 18p |
| $1000 \mu \mathrm{~F}$ | 11p | $6.8 \mu \mathrm{~F}$ | $6 \pm 1$ | $330 \mu \mathrm{~F}$ | 22p |
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| ACY17 | 24p | BCl47 | 8 p | BF178 | 20p |
| AUY18 | 21p | BC148 | 9p | BF179 | 35p |
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| ACY20 | 22p | [SC153 | 18p | BF195 | 120 |
| ACY21 | 23p | BC154 | 17p | BF444 | 87D |
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| AD142 | 44p | BC167 | 13p | BF390 | 87p |
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| AF114 | 18p | BC183L | 11p | BFY50 | 210 |
| AF115 | 18p | BC184L | 110 | BFY51 | 17p |
| AFl16 | 18p | BC186 | 330 | BFY52 | 17p |
| AF117 | 18p | BC212L | 110 | BFY 64 | 39 p |
| A F118 | 82p | BC213L | 11p | BFY90 | 72p |
| AF124 | 27p | BCL14L |  | 18X20 | 19p |
| AF139 | 39p | BC258 | 9 p | C 407 | 22p |
| A $\mathrm{F}^{\prime} 239$ | 41p | BCO59 | 9 p | C426 | 33D |
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| ALl02 | 880 | BC268 | 150 | C480 | 17p |
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| BC108 | 11p | BC302 | 30p | MP8113 | p |

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10 Silicon npn power transistors (2N 3055 ), tented/unmarked. 30 Plastic FET S, unmarked/unteated
Sinilar to 2N3819.
20 TO 5 transistors npn 2 to 5A, usitester funmarked. 20 TO 18 transistors pnp like BCl 78 . BC179, etc., untested/unmarked. 30 Plastic $2 N 3055$, unmarked/un tested. TO220 case. 10 Gieneral purpose, fully tested
500 Carbon resistors,
500 Carbon resistors, ${ }^{1}$,
100 Electrolytic contlensera
250 Ceramic Polystyrene, silver mica, 250 Ceramic Fol
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25 Potentioneters, assorted.
$2 \overline{5} 0$ High Stab, $1 \%$. $2 \%$, $5 \%$, resistors.
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Any $\overline{0}$ Packe \&4. P. \& P. 10p for $^{2}$ ach pack

Transformers
Primary 200-220-240V

|  |  |  | $\varepsilon$ | P./P. |
| :---: | :---: | :---: | :---: | :---: |
| MT111 | 12 V | $0 \cdot 5 \mathrm{~A}$ | 1-11 | 16p |
| MT111/3 | 24 V | 0.25 A | $1 \cdot 35$ | 20p |
| MT] 1 | 12 V | 2 A | 1.78 | 22p |
| MT71/3 | 245 | 1A | 1.76 | 22p |
| MT18 | 12V | 4A | $2 \cdot 47$ | 361 , |
| MT18/8 | 24 V | 2A | 2.47 | 36p |
| 30 V transformers |  |  |  |  |
| Prim 200-220-240v. Secondary voltage, |  |  |  |  |
| $12,15,20,24,30 \mathrm{~V}$ or $12-0-12 \mathrm{~V}$ or $15-0-16 \mathrm{~V}$. |  |  |  |  |
|  |  |  |  |  |
| MT112 | 0.0 A |  | $1 \cdot 32$ | 22p |
| MT79 | 1 A |  | 1.80 | 36p |
| MT3 | $\cdots \mathrm{A}$ |  | 2.98 | 36p |
| MT20 | 3A |  | $3 \cdot 30$ | 42p |
| MT51 | 5 A |  | 4.84 | 49p |

50V transformers
Prim. 200-220-240V. Secondary voltage, $19,25,33,40,50 \mathrm{~V}$ or $25-0-25 . \quad \mathbf{E} . / \mathbf{P}$.

| MT102 | 0.5 A | 1.7 C | P./P. |
| :--- | :--- | ---: | ---: |
| MT103 | 1 A | $\mathbf{8 . 5 8}$ | 38 p |
| MT105 | 3 A | $\mathbf{4 . 8 4}$ | 52 p |
| MT106 | 4 A | $\mathbf{8 . 0 2}$ | 22 p |

60V transformers
Prim. $200-220-240 \mathrm{~V}$. Secondary voltage, $24,30,40,48,60 \mathrm{~V}$ or $24-0-24 \mathrm{~V}$ or $30-0-30 \mathrm{~V}$.
MT124AT 0.5 A MT126A
MT127
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Ref.
No.
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18
70
108
72
116
1
11
18
226

20
21
51
117
88
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[^3]
# R 



# Remember! one of the Top Ten Accessory Awards from Motor magazine PUSH BUTTON CAR RADIO KIT 

NOTE: The ability to solder on a printed circuitboard is necessary to complete this kit successfully. Circuit diagram and comprehensive instructions 55 p.free with kit.

## Car Radio Kit

$\mathbf{f 6 . 6 0}+\mathbf{5 5 p}$. postage 8 packing.
Speaker including baffle and fixing strips
$\mathbf{£ 1 . 6 5 + 2 3 p . p o s t a g e ~ \& ~ p a c k i n g . ~}$
Recommended Car Aerial - fully retractable and locking.
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## stre9



QUALITY SOUND FOR LESS THAN $£ 19 \cdot 00$
Stereo 21 easy to assemble audio system kit. - no soldering required Includes:-
BSR 3 speed deck, automatic, manual facilities together with ceramic cartidge.
Two 8" 5" speakers with cabinets.
Amplitier module. Ready built with control panel, speaker leads and full, easy to follow assembly instructions.

For the technically minded :-
Specifications
Input sensitivity 600 mV : Aux. input sensitivity 120 mV : Power output 2.7 watts per channel: Output impedance $8-15$ ohms. Stezeo headphone socket with automatic speaker cutout. Provision for auxiliary inputs - radio, tape, etc., and outputs for taping discs. Overall Dimensions. Speakers approx.
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Specially selected paiz of stereo headphones with individua. level controls and padded eaf pieces to give optimum performance. $\mathbf{£ 3 . 8 5}$.

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Elegan: selt selector push button plaver for use will your own stereo system. Compathle with Viscount III system. the Stereo 21 and the Unisound module.
Technical specification
Mans input. 240V. Output sensitivity 125 mV Comparabie unit sold elsewhere at $£ 24.00$ approx.
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For the man who wants to design his own stereo -here's your chance to start, with Unisound-pre-amp. power amplifier and control panel. No soldering-just simply sctew together 4 watts per channel into 8 ohms. Inputs: 120 mV (for ceramic cattridge). The heart of Unisound is high efficiency I.C. monolithic power chips which ensure very low distortion over the audio spectrum
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Reliant Mk IV Mono Amplifier, ideal for the small disco or house parties.
Outputs 20 watts R.M.S. into 8 ohms (suitable for 150 hms ).
Inputs *5 Electrically Mixed Inputs. *3 Individual Mixing controls. *Separate bass and treble controls common to all 5 inputs.
*Mixer employing F.E.T. (Field Effect Transistors). *Solid State Circuitry. *Attractive Styling.

1) Crystal Mic of Guitar 9 mV . 2) Moving coil Mic of Guitar 8 mV . 3), 4), 5) Medium output equipment (Gram. Tuner, Monitor, Organ, etc.) all 250 mV sensitivity.

Size approx. 12 ins. $\times 6$ ins. $\times 3$ ins $\mathbf{f 1 3 . 5 0}+60 p$. postage $\&$ packing.
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A suitable 3 speed tape deck, less heads. Caters up to $5 \frac{3}{4}$ ins. spools. 240 V AC mains. Unused but store soiled hence no warranty. $\mathbf{£ 4 . 0 0}+\mathbf{£ 1 . 0 0 p \& p}$


[^4]\mathrm{ watt carbon
Capacitors
C1 0.01\muF ceramic 100V
C2 10\muF elect. 25V
C3-C6 0.01 }\textrm{F}\mathrm{ polyester (4 off)
Integrated Circuits

| IC1-IC3 | SN7473 (3 off) |
| :--- | :--- |
| IC4 | SN7400 |
| IC5 | SN7420 |
| IC6 | SN7447 |
| IC7-IC8 | SN7489 (2 off) |
| IC9-IC10 | SN7402 (2 off) |

```

Transistors
TR1-TR2 BC108

\section*{Switches}

S1-S2 Single pole change over, break before make push-button.
S3 Single pole change over miniature toggle
S4 Single pole change over, break before make push-button
S5-S12 Single pole change over miniature toggle (8 off)
S13-S16 Bank of four, four pole change over switches
All switches available from Home Radio
Potentiometer
VR1 250k \(\Omega\) carbon lin.
Lamp
LP1 Minitron 3015F 7 segment indicator
Miscellaneous
Contil Mod 2B case (West Hyde Developments) 2 plug in Veroboards \(5.1 \mathrm{in} \times 3.744 \mathrm{in}\), 0.1 in matrix to suit 32 contact edge connectors, type B

\section*{POWER SUPPLY UNIT}

Resistors
R5 \(7 \Omega 2 \mathrm{~W}\)
R6 \(270 \Omega \frac{1}{2} W\)
Capacitors
C7 \(2,000 \mu\) F elect. 25 V
C8 \(1,000 \mu \mathrm{~F}\) elect. 15 V
Transistors
TR3 2N3054
Diodes
D1 ZL4.7 4.7V Zener 1 W
D2 ZL18 18V Zener 500 mW
Bridge Rectifiers
D3-D6 MDA 942A1 Motorola
D7-D10 MDA 942A1 Motorola
Transformer
T1 Pri-240V, Sec. 1 18V, 0.35A; Sec. 26.3 V 0.7A (Henry's Radio)

Miscellaneous
FS1-200m/A cartridge fuse S17-Double pole main toggle switch
\begin{tabular}{|c|c|c|}
\hline & \multicolumn{2}{|l|}{BASS DRUM} \\
\hline \multicolumn{3}{|l|}{Resistors} \\
\hline R7 & 4.7k \(\Omega\) R13-R14 & \(56 \mathrm{k} \Omega\) (2 off) \\
\hline R8 & \(1 \mathrm{M} \Omega\) R15 & \(1.2 \mathrm{M} \Omega\) \\
\hline R9 & \(2.7 \mathrm{k} \Omega \mathrm{R} 16\) & \(100 \Omega\) \\
\hline R10 & 82k \(\Omega\), R17 & \(68 \mathrm{k} \Omega\) \\
\hline R11 & 2.2k \(\Omega\) R18 & \(10 \mathrm{k} \Omega\) \\
\hline R12 & \(2 \cdot 7 \mathrm{k} \Omega\) & \\
\hline All 10\% \(\frac{1}{2}\) & W carbon & \\
\hline
\end{tabular}

\section*{Capacitors}
\begin{tabular}{ll} 
C9 & \(0.1 \mu \mathrm{~F}\) \\
C10 & \(0.02 \mu \mathrm{~F}\) \\
C11-C15 & \(0.1 \mu \mathrm{~F}(5\) off \()\)
\end{tabular}

\section*{Potentiometers}

VR3 \(5 \mathrm{k} \Omega\) miniature preset
VR4 \(50 \mathrm{k} \Omega\) miniature preset
Transistors
TR4-TR5
BC109

\section*{HIGH AND LOW BONGOS}
\begin{tabular}{rl} 
Resistors & \\
R19-R20 & \(4.7 \mathrm{k} \Omega\) (2 off) \\
R21-R22 & \(2.7 \mathrm{k} \Omega\) (2 off) \\
R23-R24 & \(1 \mathrm{M} \Omega\) (2 off) \\
R25-R26 & \(27 \mathrm{k} \Omega\) (2 off) \\
R27-R28 & \(2 \cdot 2 \mathrm{k} \Omega\) (2 off) \\
R29-R30 & \(6.8 \mathrm{k} \Omega\) (2 off) \\
R31-R34 & \(56 \mathrm{k} \Omega\) (4 off) \\
R35-R36 & \(1.2 \mathrm{M} \Omega\) (2 off) \\
R37-R38 & \(68 \mathrm{k} \Omega\) (2 off) \\
R39-R40 & \(100 \Omega\) (2 off) \\
R41-R44 & \(82 \mathrm{k} \Omega\) (4 off) \\
All \(\frac{1}{2} W\) & \(10 \%\) carbon \\
Capacitors & \\
C16-C17 & \(0.01 \mu \mathrm{~F}\) (2 off) \\
C18-C23 & \(0.015 \mu \mathrm{~F}\) (6 off) \\
C24* & \(0.033 \mu \mathrm{~F}\) (High bongo only) \\
C25* & \(0.1 \mu \mathrm{~F}\) (Low bongo only) \\
C26-C27 & \(0.05 \mu \mathrm{~F}\) (2 off) \\
C28-C29 & \(0.01 \mu \mathrm{~F}\) (2 off)
\end{tabular}

Transistors
TR6-TR9 BC109 (4 off)
Potentiometers
VR5-VR6 \(5 \mathrm{k} \Omega\) miniature preset (2 off)
VR7-VR8 \(50 \mathrm{k} \Omega\) miniature preset (2 off)
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|c|}{BLOCKS} \\
\hline \multicolumn{4}{|l|}{Resistors} \\
\hline R45 & \(4.7 \mathrm{k} \Omega\) & R50 & \(68 \mathrm{k} \Omega\) \\
\hline R46-R47 & \(27 \mathrm{k} \Omega\) (2 off) & R51-R52 & \(22 \mathrm{k} \Omega\) \\
\hline R48 & 150k \(\Omega\) & R53 & 15k \(\Omega\) \\
\hline R49 & \(390 \mathrm{k} \Omega\) & R54 & \(2.7 \mathrm{k} \Omega\) \\
\hline All \(10 \% \frac{1}{2}\) & watt carbon & & \\
\hline \multicolumn{4}{|l|}{Capacitors} \\
\hline C30 & \(0.1 \mu \mathrm{~F}\) & & \\
\hline C31-C33 & \(0.002 \mu \mathrm{~F}\) (3 & & \\
\hline C34 & \(0.1 \mu \mathrm{~F}\) & & \\
\hline C35 & 500 pF & & \\
\hline
\end{tabular}

Potentiometers
VR9 \(50 \mathrm{k} \Omega\) miniature preset
Transistors
TR10-TR12 BC109 (3 off)

\section*{SNARE DRUM}

Resistors
\begin{tabular}{llll} 
R55-R56 & \(4.7 \mathrm{k} \Omega\) & R61 & \(12 \mathrm{k} \Omega\) \\
R57 & \(3.3 \mathrm{k} \Omega\) & R62 & \(120 \mathrm{k} \Omega\) \\
R58 & \(1 \mathrm{k} \Omega\) & R63 & \(2.2 \mathrm{k} \Omega\) \\
R59 & \(120 \mathrm{k} \Omega\) & R64 & \(3.9 \mathrm{k} \Omega\) \\
R60 & \(100 \mathrm{k} \Omega\) & R65 & \(15 \mathrm{k} \Omega\) \\
All \(\frac{1}{2}\) watt \(10 \%\) carbon &
\end{tabular}

\section*{Capacitors}

C36 \(1 \mu \mathrm{~F}\) elect. 15 V
C37 \(10 \mu \mathrm{~F}\) elect. 15 V
C38 \(0.1 \mu \mathrm{~F}\)
C39 \(0.0022 \mu \mathrm{~F}\)
C40 \(0.022 \mu \mathrm{~F}\)
C41 220 pF
C42 \(1.5 \mu \mathrm{~F}\) elect. 15 V
C43 500 pF
Potentiometer
VR10 \(100 \mathrm{k} \Omega\) miniature preset
Transistors
TR13-TR14 BC109 (2 off)
Diodes
D11 1N914
D12 Z1J (Semitron)
Inductor
L1 OL130 transformer
(Henry's Radio)

\section*{CYMBAL}

Resistors
\begin{tabular}{llll} 
R66 & \(3.3 \mathrm{k} \Omega\) & R73 & \(120 \mathrm{k} \Omega\) \\
R67 & \(10 \mathrm{k} \Omega\) & R74 & \(100 \mathrm{k} \Omega\) \\
R68 & \(470 \mathrm{k} \Omega\) & R75 & \(120 \mathrm{k} \Omega\) \\
R69 & \(47 \mathrm{k} \Omega\) & R76-R77 & \(10 \mathrm{k} \Omega\) \\
R70 & \(10 \mathrm{k} \Omega\) & R78 & \(2.2 \mathrm{k} \Omega\) \\
R71 & \(1.5 \mathrm{M} \Omega\) & R79 & \(1 \mathrm{M} \Omega\) \\
R72 & \(680 \Omega\) & R80 & \(10 \mathrm{k} \Omega\)
\end{tabular}

All \(\frac{1}{2}\) watt \(10 \%\) carbon
\(\begin{array}{lll}\text { Capacitors } & \mathrm{C} 47 & 0.05 \mu \mathrm{~F} \\ \mathrm{C} 44 & 2 \mu \mathrm{~F} \text { elect. } 15 \mathrm{~V} & \mathrm{C} 48 \\ & 1 \mu \mathrm{~F} \text { elect. } \\ \mathrm{C} 45 & 8 \mu \mathrm{~F} \text { elect. } 15 \mathrm{~V} & \\ \mathrm{C} 46 & 15 \mathrm{~V} \\ & 0.1 \mu \mathrm{~F} & \\ \mathrm{C} 49 & 0.05 \mu \mathrm{~F}\end{array}\)
Transistors
TR15-TR17 BC109 (3 off)
Diodes
```

D13-D14 1N914 (2 off)
D15 Z1J (Semitron)

```

Inductor
L2 0 O130 transformer
LONG BRUSH
\begin{tabular}{|c|c|c|c|}
\hline \multicolumn{4}{|l|}{Resistors} \\
\hline R81 & \(4.7 \mathrm{k} \Omega\) & R88 & , \\
\hline R82-R83 & R83 22k \(\Omega\) & R89-R90 & 10k \\
\hline R84 & 1.2M \(\Omega\) & R91 & \(2 \cdot 2 \mathrm{k} \Omega\) \\
\hline R85 & \(390 \Omega\) & R92 & \(150 \mathrm{k} \Omega\) \\
\hline R86 & 100k \(\Omega\) & R93 & 82k \(\Omega\) \\
\hline R87 & \(68 \mathrm{k} \Omega\) & & \\
\hline \multicolumn{4}{|l|}{All \(\frac{1}{2}\) watt 10\% carbon} \\
\hline Capacito & itors & C53 & 0.01 \\
\hline C50 \(4 \mu\) & \(4 \mu \mathrm{~F}\) elect & V C54 & \(2.5 \mu \mathrm{~F}\) ele \\
\hline C51 0.1 & \(0.1 \mu \mathrm{~F}\) & & 15 V \\
\hline C52 0.02 & \(0.02 \mu \mathrm{~F}\) & C55 & \(0.22 \mu \mathrm{~F}\) \\
\hline \multicolumn{4}{|l|}{Transistors} \\
\hline Diodes
\[
\text { D16 } 1 \mathrm{~N}
\] & \[
{ }^{s} \text { N914 }
\] & 7 Z1J ( & (Semitron) \\
\hline
\end{tabular}

SHORT BRUSH

Resistors
\begin{tabular}{ll} 
R94 & \(4.7 \mathrm{k} \Omega\) \\
R95-R96 & \(22 \mathrm{k} \Omega 2\) (2 off) \\
R97 & \(1.2 \mathrm{M} \Omega\) \\
R98 & \(390 \Omega \Omega\) \\
R99 & \(100 \mathrm{k} \Omega\) \\
R100 & \(68 \mathrm{k} \Omega\) \\
R101 & \(100 \mathrm{k} \Omega\) \\
R102-R103 & \(10 \mathrm{k} \Omega(2\) off \()\) \\
R104 & \(2.2 \mathrm{k} \Omega\) \\
R105 & \(150 \mathrm{k} \Omega\) \\
R106 & \(82 \mathrm{k} \Omega\) \\
All \(\frac{1}{2}\) watt \(10 \%\) carbon
\end{tabular}

\section*{Capacitors}

C56 \(1 \mu \mathrm{~F}\) elect. 15 V
C57 \(0.1 \mu \mathrm{~F}\)
C58 \(0.02 \mu \mathrm{~F}\)
C59 \(0.01 \mu \mathrm{~F}\)
C60 \(2.5 \mu \mathrm{~F}\) elect. 15 v
C61 \(0.22 \mu \mathrm{~F}\)
Transistors
TR21-TR23 BC109 (3 off)
Diodes
D16 1N914
D17 Z1J (Semitron)
MIXER AND PRE-AMPLIFIERS
Resistors
\begin{tabular}{ll} 
Relo7 & \(4.7 \mathrm{k} \Omega\) \\
R108-R112 & \(2.2 \mathrm{k} \Omega\) (5off) \\
R113 & \(10 \mathrm{k} \Omega\) \\
R114 & \(2 \mathrm{M} \Omega\) \\
R115 & \(10 \mathrm{k} \Omega\) \\
R116 & \(2.9 \mathrm{k} \Omega\) \\
All \(\frac{1}{2}\) watt \(10 \%\) carbon
\end{tabular}

\section*{Capacitors}

C62-C66 \(0.01 \mu \mathrm{~F}\) (5 off)
\(\mathrm{C} 67 \quad 0.05 \mu \mathrm{~F}\)
Potentiometers
VR11-VR15 \(2 \cdot 5 \mathrm{M} \Omega\) miniature preset (5 off)

Transistors
TR24-TR29 2N3819 (6 off)

\section*{AMPLIFIER}

Resistors
\begin{tabular}{|c|c|c|c|}
\hline R117 & 330k \(\Omega\) & R121 & 39k \(\Omega\) \\
\hline R118 & 39k S & R122 & 3.3k \(\Omega\) \\
\hline R119 & 18k S & R123 & \(22 \Omega\) \\
\hline R120 & 150k \(\Omega\) & R124 & 1 k S \\
\hline \multicolumn{4}{|l|}{All \(\frac{1}{2}\) watt 10\% carbon} \\
\hline \multicolumn{4}{|l|}{Capacitors} \\
\hline C68 & \(0.1 \mu \mathrm{~F}\) & C71 \(0.1 \mu \mathrm{~F}\) & \\
\hline C69 & \(0.002 \mu \mathrm{~F}\) & C72 \(50 \mu \mathrm{~F}\) el & ect. 15 V \\
\hline C70 & \(500 \mu \mathrm{~F}\) el & ct. 15 V & \\
\hline
\end{tabular}

Potentiometer VR16 25k \(\Omega\)

Integrated Circuit
IC1 PA263 (Jermyn)

\section*{Loudspeaker}

LS1 \(8 \Omega, 5\) watt 4in loudspeaker


Fig. 2. Circuit of tempo generator

Fig. 3. Circuit of divide by four, four bit counter, time signature decoder and step generator


\section*{TEMPO GENERATOR}

The circuit for the tempo generator is shown in Fig. 2. If the multivibrator capacitors were of equal value, the output would be a symmetrical square wave but C 1 is \(1 / 1000\) the value of C 2 creating a pulse of very short duration at a repetition rate determined largely by the value of C 2 . The output period is made variable by VR1 and gives a sufficient range of tempo to cover most requirements.

The output feeds a divide by four counter stage (IC1) composed of flip-flops A1 and A2 as shown in Fig. 3. The output of this counter is a symmetrical square wave with a period variable from 2 seconds to 25 milliseconds. This output is the basic timing waveform for the entire system.

\section*{FOUR BIT COUNTER}

The four flip-flops labelled B1, B2, C1 and C2 (IC2, IC3) make up a four bit binary counter. Drive for this counter is obtained either from the output of the divider in the "Run" position of S3, the "Stop/Run" switch, or from the bistable latch (D1 and D2) in the "Stop" position. This allows the counter to be stepped through the count sequence manually with S1, the "Step" switch.

All four flip-flops can be reset by grounding the reset pins ( 2 and 6 of each chip). To reset the counter in the "Stop" position, the "Reset" switch S4 is pressed. This forces the counter to binary 0000.

In the \(4 / 4\) position of the "Time Signature" switch S13, the counter will count through 16 counts and automatically return to 0000 . In the \(3 / 4\) and \(6 / 8\) positions, the counter must be forced back to 0000 after 12 counts.

This is done by applying the \(\overline{\mathrm{Q}}\) outputs of B1 and B2 and the Q outputs of Cl and C 2 to the input of NAND gate E1 (IC5). When this count is reached, the output of that gate will go immediately to 0 and the counter will be reset.

For \(9 / 8\) time, only 9 counts are allowed and the counter is then reset with Nand gate E2.


Prototype p.c.b. including tempo, timing and logic assembly. Four \(0.01 \mu \mathrm{~F}\) capacitors must be included for decoupling the chip supply lines

\section*{STORAGE, OUTPUT AND INDICATOR}

Fig. 4 shows the logic diagram for the storage, indicator and output gates. The four counter Q outputs from IC2/IC3 are applied to the A, B, C and \(D\) inputs of each of the memory chip address select lines.

S5 through S12 are the instrument input switches and S2 is the "Write" switch. The chip select inputs (pins 2) on both IC7 and IC8 are permanently grounded which simply means that the output will always be available.

The output of these storage or memory chips is applied to the inputs of eight NOR gates (IC9/IC10) and gated with the output of the divider IC1 to give a positive pulse out for each position and time segment that an instrument is to sound.

IC6 uses outputs A, B and C of the counter to decode the correct sequence of grounds to be applied to the Minitron indicator.

\section*{LOGIC ASSEMBLY}

Although the prototype used plug-in printed circuit boards, 0.1 in plug-in Veroboards can be successfully substituted for the logic and percussion circuitry.

While no constructional details are included sufficient information for this should be gleaned from the photograph of the prototype logic card showing the disposition of the tempo generator and logic chips. Note carefully the use of four \(0.01 \mu \mathrm{~F}\) capacitors connected between the supply and ground lines. These should be added to prevent spurious triggering.

Point to point wiring on the board should be carried out according to the circuit diagrams given. It really boils down to wiring between the i.c. pin numbers shown and taking off external wiring, such as to control switches and percussion generators, to those copper strips which will mate with the Veroboard's edge connector.


Fig. 4. Circuit for storage, lamp driver and percussion gates


Fig. 5. Circuit of power unit


AMPLIFIER

Part of the component assembly within the case. The amplifier is outboard mounted and the speaker connected to the removed plate

Once connections have been made and verified breaks in the copper should be made where appropriate. Finally check all wiring according to the circuits and make sure that there are no solder bridges or loose bits of wire around to cause future trouble.

\section*{POWER SUPPLY}

The power supply is made up on a \(2 \frac{1}{2}\) in \(\times 2 \frac{1}{2}\) in, 0.15 in matrix Veroboard. With the transformer and large electrolytic C7 these are all finally mounted on the Contil case chassis plate. With the wiring completed for this connect to the mains and ensure that the line output voltages are correct.

At this point it is a good idea to cut and drill those holes for the control panel switches and indicator. The locations for these are indicated in the photograph.

The 4.7 V power lines can now be connected to the logic board edge connector and the appropriate connections to the control panel switches made. A number of tests can now be made.

First plug in the board and switch on. In the "Stop" position of S3, press the "Reset" button. The Minitron indicator should show a figure eight. This is a lamp test feature and assures that all seven segments of the indicator work correctly. With the switch released the indicator should read 0. With S3 now in the "Run" position the Minitron indicator should go through 0 to 7 repetitively. The "Tempo" control can slow this process right down so that single digit changes can be easily observed.

If these tests are satisfactory, the unit is sound to this point. There are other tests that could be made but they would require additional instruments and are not really worth including.

Next month details for the percussion generators, mixer and amplifier will be given.


The Public Address Engineers' Exhibition, Bloomsbury Centre Hotel, Russell Square, London WC1 12-14th March, 1974.
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superb solid state audio amplifer. Brand throughout. jo silicon transistors plus "
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 Uaing heavy duty fully mains transmains trans-
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ade quate moothing
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adjusted by means of a separate "balance" control fitted at the rear of the chassis. Input senstivity is approximately \(300 \mathrm{~m} / \mathrm{v}\) for full peak output of 4 watts per chammel ( 8 watta mono). into 3 ohm 日peakers. Full negative feedback in a carefully calculated circuit, nllows high volump levels to be used with negligilile distortion. Supplied complete with knobs, chassis size \(11 \mathrm{in} . w \times 4 \mathrm{In}\). \(x\). Orerall height including valver sir. Ready built and
tested to a high standard. Price \(\mathbf{9 . 9 0 . ~ P . ~ \& ~ P . ~ 5 0 p . ~}\)

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 outputs as on Joad.
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\hline & & & & & \multicolumn{5}{|l|}{Log or lin less switch} \\
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\end{tabular}}} & \multicolumn{4}{|l|}{Dual log less switch 250K,} & D \\
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\section*{FLUORESCENT}

\section*{LAMP INVERTER}

\section*{High efficiency light source from a 12 V d.c. supply}

WITH power cuts an annual recurrence the construction of an inverter to drive a fluorescent lamp from a 12 V d.c. supply, such as a car battery, is a worthwhile investment.

Although one basic circuit is described, with minor component changes the inverter is capable of driving either an \(8 \mathrm{~W}, 13 \mathrm{~W}\) or 20 W lamp.

\section*{HIGH FREQUENCY OPERATION}

The choice of an operating frequency above 20 kHz for the inverter means a reduction in transformer size, increased lamp efficiency and the elimination of audible whistles.

The unit incorporates transistor protection against lamp removal by automatically stopping the oscillator. Diode protection against supply polarity reversal is also included.

\section*{CIRCUIT DESCRIPTION}

The circuit shown in Fig. 1 is developed from the well-known Hartley oscillator. The primary winding of the pot core transformer TI is in the emitter circuit so that the collector of the transistor TR1 can be earthed. This arrangement means that the feedback winding must give a greater voltage than the primary winding and therefore has more turns.

The feedback winding provides the drive to the transistor base via a limiting resistor R1, one of the lamp heaters and C3.
The voltage across the secondary winding is sufficient to strike the tube and the leakage inductance in conjunction with the base resistor limits the tube running current. This is necessary because the fluorescent tube tends to maintain a constant voltage.
Heater current for the tube is provided at one end by the drive current and at the other by a few extra turns on the secondary winding. The bypass capacitor across the transistor limits the peak collector emitter voltage to a safe value. This is not normally required in the 8 W version.

\section*{POSITIVE AND NEGATIVE EARTH}

If the circuit is being constructed in a separate metal case and used with a positive earthed supply system the transistor can be mounted directly on to the case. If the supply is negative earthed a mica washer and plastic bushes should be used to insulate the transistor which should be protected from possible shorts to earth by a plastic cover.
The circuit board should now be cut and drilled to fit the case and the components mounted as shown. The transformer leads can be connected to the printed circuit via pins and further pins can


Fig. 1. Circuit diagram of fluorescent lamp inverter
be used for connecting leads to the transistor and the fluorescent lamp.
When the wiring-up is completed the circuit should be carefully checked before it is connected to the voltage supply to ensure that both diodes and the transistor are connected the right way round and that copper breaks in the Veroboard have been cleanly cut and that there are no shorts between strips due to surplus solder.

\section*{TESTING}

Testing should start with a low voltage battery or a gradually increasing voltage supply with a series ammeter to check that the current consumption is not excessive and that the transistor is not overheating.

The fluorescent tube must be connected to complete the bias circuit and if the circuit is oscillating a glow will be seen from the heaters at the ends of the tube. When the full voltage is applied the lamp should strike and light up but during preliminary testing the lamp will strike easier if a fine wire is stretched between the metal end caps and connected to one of the heater pins.

If the circuit does not oscillate L1 may be reversed or there may be some shorted turns in the transformer. When operating normally with a 13 watt tube this inverter consumes about 1.4 amps at 12 volts. This current and the light output can be laried slightly by small adjustments to R1 and/or C3.

\section*{PRECAUTIONS}

All metal parts, including the metal end caps on the tube, should be earthed and the fine wire connected to the heater pin during tests should be removed. The voltage across the tube is high enough to provide a nasty shock and should be treated with respect.

\section*{TRANSFORMER WINDING}

The winding order is identical for the 8 watt, 13 watt or 20 watt versions of the inverter. Start by winding the primary Ll on the bobbin according to the relevant data given in the Components List. Lead ends on each coil should be colour coded for later identification.

Insulate with plastic adhesive tape and then add L2 again insulating the winding with tape. Finally


\section*{Interior assembly of 13W inverter}
add L3 and L4. The direction of winding for all coils is the same.

Before bolting the two halves of the Ferroxcube pot core together onto the Veroboard, two paper spacing washers should be cut and fitted between them.

\section*{CONSTRUCTION}

The Ferroxcube transformer and majority of circuit components are mounted on a \(2 \mathrm{in} \times 2 \frac{1}{2} \mathrm{in}\), \(0 \cdot 1\) in matrix Veroboard with the transistor connected to case.

The mounting of the fluorescent lamp will depend on the application and to a certain extent on the constructor's preferences. For example an 8 watt tube can be used in a caravan mounted flush with the roof and components separate or on a channel support with the components mounted in the channel.
Alternatively an attractive table or hanging lamp can be constructed with the fluorescent lamp in a vertical Perspex tube with the components in a suitable case at one end of the tube.
Another idea is to construct a portable lamp using two 6 V lantern batteries in a suitable case with the lamp and reflector mounted on the outside. Leads can then be connected via a plug and socket for external battery supplies such as a car battery: This makes an ideal camping light but the lantern batteries should be isolated from the car battery by diodes or a switch.

The 13W inverter with output leads terminated with clips and lamp


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\section*{SKYLAB-LAST MISSION}

As this is being written the last Skylab mission planned before the Soviet-American rendezvous in space, is not quite half way through. The longest single mission of three astronauts will bring to an end the triple series of scientific programmes.

By. whatever standards these experiments may be judged they experiment one of the major events in the history of the solar system. Some may regard the Skylab project as more important than the Apollo programme. Certainly, a very comprehensive coverage of observation (and running repairs) will add much to the data so far amassed. It must not, however, be forgotten that side by side with the success story go some rather disquieting happenings.

It is intended that the return of the team will be in early February. However, at the time of writing, there are possibilities that there could be an earlier termination because of certain disciplinary reactions. Perhaps it is to be a blessing, if seen in the broad context of advanced technology.

The problems that have arisen are twofold. There have been technical difficulties and some personnel problems. None of the present crew have flown in space before and this could to some extent explain the behavioural problems.

\section*{CREW CONCERN}

Mission Control have been concerned about the crew in that they suffered from motion sickness for several days and attempted to conceal this from the physicians monitoring the flight. Perhaps this is understandable since they would be worried about being recalled. On the other hand there were other signs such as proneness to error and poor disposition.
Their sleep sessions have tended to be longer than was thought to be required, compared with the previous crews. They have been moody and temperamental when in contact with control and have complained somewhat bitterly about their workload. Refusal to come in from extra vehicular activities has seemed to show a somewhat antagonistic attitude towards instructions.

The consequences on their work has been very noticeable. The most serious error has been in relation to the operation of the cameras during observation passes of the Earth. The proper filters to six cameras were forgotten. Much of the film will be useless though some may be recoverable using certain


BY FRANK W. HYDE
special techniques when they are returned to base.

The medical unit in charge at control will be seeking an answer to these rather irrational episodes. Despite this however, it cannot detract from the remarkable Skylab project.

Some of the technical problems have been concerned with the gyroscopes. One of these has been out of use since the end of November. The other has tended to show a slowing down. Normally it would have a revolution of 8,200 per minute. It slowed by 14 revolutions and then by a further 10 per minute. This is quite serious and should it continue to fall then the mission could well be curtailed.

\section*{PIONEER 10 AND JUPITER}

Some of the vital statistics so far recorded during Pioneer's mission to Jupiter have confirmed previous thinking. The magnetic field for example is recorded at 4 gauss. In a recent "SPACEWATCH" mention was made that Dr C. H. Barrow had used a figure of 9 to 10 gauss. Re -examination of his previous papers can now lead to some constructive theories.

The off-set magnetic field confirms Warwick's claim made several years ago. The higher level of the radiation belts has also been confirmed. It is also confirmed that lo has an ionosphere. This supports the earlier data that Io has an influence on the decametre radiation from Jupiter.

It will be some time yet before the accumulated data is to hand for distribution.

Perhaps another important point that might be made is the fact that so far the 8 W transmitter has ful-
filled its task admirably. It remains to be seen (heard?) how far out it will continue to supply data.

One further important development is that it is now quite definitely confirmed that there are hazards to control, in the magnetosphere of Jupiter. There is the possibility of random effects giving the same sort of command signal to the vehicle.

The case of far out probes such as Pioneer 10 , introduces the long time period of signal travel. That is 45 minutes out and 45 minutes back. This could be much too long to counteract false signals. To overcome this a method of repeating commands every 30 minutes is being used. By this means the risks are minimised.

\section*{MOON AS A STANDARD LAMP}

Dr J. Linsky of Boulder. Colorado, has made use of data obtained from the Apollo missions and the study of the lunar soil, to set the moon as a standard of comparison for infra-red and microwaves.

The average values have been studied and Linsky's results have shown an accuracy of 4 per cent for wavelengths between 10 micrometres and 1 metre. By averaging the radiation between successive new moons he arrived at a figure for radiation over a lunar day for the central part of the moon's disc. It is possible to use this standard to calibrate the sun and parts of the galaxy.

Linsky is able to fix the brightness at an order of magnitude better than with other methods. Possibilities of a better study of the chromosphere are therefore indicated and perhaps a better view of how the heat transfer shock waves operate.

\section*{QUASARS}

The see-saw of opinion continues to operate in the controversial field of quasars. In spite of this there is growing confirmation of the claims of the radio astronomers that these are far distant rapidly receding objects.

Recent new work has offered a new dilemma for theorists since in one particular case the observers are faced with deciding as to whether, in certain cases, the quasar is in fact way beyond its apparent position relative to a galaxy or whether Hubbles law is in doubt in such cases.

It still leaves the position largely as before, that of all the new discoveries of other sources such as pulsars, black holes, X-ray stars, etc., the quasar remains the most mysterious. Near or far as to position still remains in doubt when looking at the overall picture.

One advantage of inserting a resistor in series with the source is that the slope resistance of the device is increased. This is because a change in channel current causes a change in the voltage across the resistor which alters the reverse bias on the gate. Obviously, for a given change in current, the voltage change is proportional to the resistance in series with the source.

\section*{HIGH SLOPE RESISTANCE}

In applications where a very high slope resistance is required, the circuit shown in Fig. 7 can be used. In this circuit, the source resistor is replaced by an f.e.t. current generator which provides a high slope resistance in series with the source of TRI.

For proper operation, the cut-off voltage \(V_{p}\) of TR1 should be greater than that of TR2. The slope resistance can be increased by one or two orders of magnitude using this technique.

The temperature coefficient of drain current depends on the value of drain current, being negative at high currents and positive at low currents. At an intermediate value of drain current \(I_{\mathrm{D}(\mathrm{o})}\), the two effects causing these changes cancel and the temperature coefficient is approximately zero. The current at which this occurs is given approximately by:
\[
I_{\mathrm{D}(0)}=\frac{0.4}{V_{\mathrm{p}}{ }^{2}} \times I_{\mathrm{DSS}}
\]

\section*{y AND \(h\) PARAMETERS}

It is often convenient to refer to the characteristics of transistors by means of \(y\) or \(h\) parameters. There are four \(h\) and four \(y\) parameters for each of the three basic transistor connections.

The four \(y\) parameters describing a transistor operating in the common emitter configuration are: \(y_{\text {te, }} y_{\mathrm{fe}}, y_{\mathrm{re}}\) and \(y_{o c}\). The second subscript letter " \(e\) " refers to the fact that the transistor is in the common emitter connection. Similarly, "b" denotes common base, "s" denotes common source, etc. The first subscript letters refer to input, forward transfer, reverse transfer and output parameters respectively.
Each " \(y\) " parameter is the ratio of a current to a voltage. Thus:
\(y\) le is the input admittance (the ratio of input current to input voltage);
\(y\) 'se is the forward transfer admittance (the ratio of output current to input voltage);
\(y^{\prime}\) re is the reverse transfer admittance (the ratio of input current to output voltage);
\(y_{o e}\) is the output admittance (the ratio of output current to output voltage).
Whereas \(y\) parameters are all ratios of currents to voltages, different \(h\) parameters represent ratios of different quantities. Thus:
\(h_{1 \mathrm{e}}\) is the input impedance (the ratio of input voltage to input current);
\(h_{\text {fe }}\) is the forward current transfer ratio (the ratio of output current to input current);
\(h_{\text {re }}\) is the reverse voltage transfer ratio (the ratio of input voltage to output voltage);
\(h_{o e}\) is the cutput admittance.
Although common emitter parameters have been quoted, the definitions are also valid for other connections.

Of the six possible sets of parameters, only \(h\) and \(y\) parameters are widely used in transistor circuit analysis. If one set of parameters is known, any other set can be calculated quite easily. This means that the choice of parameters for a given application is largely a matter of convenience.

The characteristics of bipolar transistors are usually described by \(h\) parameters and those of field effect transistors by \(y\) parameters.

\section*{BIPOLAR TRANSISTORS}

A bipolar transistor in either the common base or common emitter configuration can also be used as a constant current source as shown in Fig. 8. Since the collector-base junction is a reverse biased diode, the current through it is only slightly affected by the voltage across it.

The slope resistance in the common base connection is \(\left(h_{\text {ife }}+1\right)\) times larger than the slope resistance in the common emitter connection. The difference between common base and common emitter circuits lies in the resistances in series with the base and emitter leads.

The circuit in Fig. 9 is a generalised common base or common emitter circuit. If \(R_{\mathrm{B}}=0\) and \(R_{\mathrm{E}}=\propto\) the slope resistance is equal to the reciprocal of the low frequency common base output admittance \(h_{o b}\). In practice, provided that \(R_{\mathrm{E}}\) is more than a few kilohms, the slope resistance is not appreciably less than this value. If \(R_{13}=\alpha\) and \(R_{\mathrm{E}}=0\), the slope resistance equals \(1 / h_{0 e}\). Then \(h_{0 e}\) and \(h_{\mathrm{rb}}\), are related by the equation:
\[
h_{\mathrm{oe}}=h_{(\mathrm{l})} \times\left(h_{\mathrm{fe}}+1\right)
\]

With intermediate values of \(R_{13}\) and \(R_{\mathrm{E}}\) the slope resistance lies between these two values. For a \(\mathrm{BC108}, h_{\mathrm{ob}}=10^{-7} \mu \mathrm{~S}\) and \(h_{o e}=3 \times 10^{-5} \mu \mathrm{~S}\).

The circuit of a simple transistor constant current generator is given in Fig. 10. The base voltage is set to 4 V by the potential divider across the supply.

Since the BC184 is a silicon transistor, the emitter voltage is about 3.4 V . Thus the emitter current is \(3 \cdot 4 \mathrm{~V} / 6 \cdot 8 \mathrm{k} \Omega=500 \mu \mathrm{~A}\) and since \(\alpha \simeq 1\), the collector

current is also about \(500 \mu \mathrm{~A}\). Because the emitter resistance is \(6.8 \mathrm{k} \Omega\), the slope resistance of this current source is very high. The knee voltage is about 4 V .

The major disadvantage of the circuit in Fig. 10 is that the current is approximately proportional to the supply voltage. To produce a more stable current source, R2 is often replaced by a Zener diode so that the base voltage, and hence the emitter voltage, is less affected by supply voltage variations.

The temperature coefficient of this source depends on the temperature coefficients of the Zener diode and the base-emitter junction. If a Zener of about 4 V is used, the temperature coefficients will cancel.

\section*{IN INTEGRATED CIRCUITS}

Fig. 11 shows the circuit of an interesting transistor current source which is often used in integrated circuits. In this circuit, the base current for the transistor is supplied by part of the current through the resistor. The remainder of the current from the resistor flows through the diode.

In an integrated circuit, since all the transistors and diodes are produced simultaneously on the same chip of silicon, the diode can be made identical to the base-emitter junction of the transistor. Because the diode and the base-emitter diode are connected in parallel, the voltages across the two are equal. Since the two diodes have the same characteristics, the currents through them ( \(I_{\mathrm{D}}\), and \(I_{\mathrm{E}}\) ) are also equal.

The current through the diode comes entirely from the resistor but the emitter current in the transistor is provided partly by the resistor and partly by the collector circuit. In other words,
\[
\begin{aligned}
I_{\mathrm{D}} & =I_{\mathrm{E}} \\
& =I_{\mathrm{C}}+I_{\mathrm{B}}
\end{aligned}
\]

The base current of the transistor is about one per cent of the emitter current and can be neglected. With this assumption, it can be seen that the collector current of the transistor is equal to the current flowing through the diode. Because the base current is small, almost all of the current flowing through the resistor
flows into the diode so that the collector current of the transistor is approximately equal to the current in the resistor.

Circuits of this type have very low knee voltages, typically less than 0.5 V . This circuit also has a very low temperature coefficient because the diode and transistor have equal temperature coefficients of voltage which cancel each other.

The breakdown voltage of a transistor current generator can be anything from 15 to 500 V depending on the transistor employed.

\section*{SOPHISTICATED CIRCUITS}

There are several more complex constant current circuits available which are used where high stability is required. The ring-of-two circuit shown in Fig. 12 is of ten used as a precision current generator. It is a two-terminal current source because, unlike the transistor circuits described above, no extra supply voltage is necessary.

The operation can be understood if it is considered as two constant current sources, the first consisting of D1, R1 and TR1 and the second containing D2, R2 and TR2. These sources are connected so that each one supplies a constant current to the Zener diode of the other. The knee voltage of the ring-of-two is rather high, and is equal to the sum of the Zener voltages of D1 and D2.
Just as it is possible to construct regulated voltage power supplies, so regulated current supplies can be made.
The current is usually measured by passing it through a resistor and comparing the voltage thus produced with a reference voltage to derive a control signal. This signal is then applied to a bipolar transistor or an f.e.t. which controls the current. A block diagram for such a scheme is shown in Fig. 13.

In the concluding part of this article next month, practical applications of constant current sources will be described.


Fig. 10. Simple transistor constant current source

Fig. 12. Ring-of-two constant current generator

Fig. 13. Stabilised current power supply



THERE are many games in which contestants' reaction times play a vital part; "snap" is an obvious example and a guessing game where the first contestant to press a button is allowed to answer is another. This simple unit has two push buttons and gives a positive indication (by lighting a lamp) of the button which is pressed first.

\section*{CIRCUIT DESCRIPTION}

The circuit of the Electronic Snap is shown in Fig. 1. It makes use of a single digital integrated circuit type SN7400 which is available for less than 20p. The indicators used are light emitting diodes which are simply diodes which emit light when forward biased.

The SN7400 contains four nand gates. A nand gate is one whose output will only go to logic 0 if both inputs are logic 1. They are cross-coupled in the circuit so as to form two bistables, circuits which


Fig. 1. Circuit diagram of the Electronic Snap

\section*{COMPDNEXTS}

Resistors
R1-R4 \(680 \Omega\) (4 off)
R5 \(150 \Omega\)
R6 \(1 \mathrm{k} \Omega\)
All \(\pm 10 \% \frac{1}{4} \mathrm{~W}\) carbon
Capacitors
C1, C2 1,000 p polystyrene (2 off)
Diodes
D1, D2 1N914 or any general purpose silicon diode (2 off)
D3, D4, D6 TIL209 or similar light emitting diode (3 off)
D5 \(\quad 5 \cdot 1 \mathrm{~V} 400 \mathrm{~mW}\) Zener diode
Integrated Circuit
IC1 SN7400N Quad 2-input NAND gate Switches

S1, S2 Push to changeover (1 pair normally closed, 1 pair normally open) (2 off)
S3 Push to changeover
S4 Miniature d.p.d.t. toggle

\section*{Miscellaneous}

B1 9V PP4 battery
0.1 in matrix Veroboard 2 in \(\times 1 \frac{1}{2}\) in

Diecast box \(4 \frac{1}{4}\) in \(\times 2 \frac{1}{4}\) in \(\times 1\) in
35 mm film cases (2 off)
3.5 mm jack plug and socket (2 off)

are stable in either of two states. When the reset button is pressed inputs to gates G1 and G3 are taken to logic 0 (less than 0.8 V ) through diodes DI and D2. The outputs of G1 and G3 must go to logic 1 regardless of the state of pins 13 and 5.

Switches S1 and S2 are open so the inputs on pin 1 of G2 and pin 10 of G4 are logic 1. Because the outputs of G1 and G3 are logic 1 it follows that all inputs to gates G2 and G4 are logic 1 so their outputs go to logic 0 . The circuit will stay in this state even when S 3 is released.

Assume S1 is now pressed. One input to G2 goes to \(\operatorname{logic} 0\), the output goes to logic 1 causing the output of Gl to go to logic 0 which, in turn, causes l.e.d. D3 to conduct and light up. Pin 5 of G3 is now held at logic 0 so even if S2 is pressed causing the output of G4 to go to logic 1 the output of G3 will stay at logic 1 keeping l.e.d. D4 off

\section*{POWER SUPPLY}

Power for the circuit is provided by a nine volt
battery which is dropped down to the five volts required by the integrated circuit by the \(150 \Omega\) resistor R5 and Zener diode D5.

One point that may be noticed is the connection of switches S1 and S2. These are not connected so as just to ground the inputs to G2 and G4 but to provide a very short pulse which is enough to take the inputs to logic 0 but will not drain the battery if the switches are held down.

The switches are connected as changeovers and the very short interval during which the capacitor charges is enough to momentarily take the output to ground.

\section*{CONSTRUCTION}

The bulk of the circuit is built on a small piece of Veroboard as shown in Fig. 2. Jack plugs are used to connect the switches to the main unit. The switches are mounted in 35 mm film cases and the capacitors Cl and C 2 are mounted on the switch contacts.


Fig. 2. Layout of the components on the Veroboard and front panel, and interwiring details

\title{
SCORPIO Mn. 2 IGNIIION System \\ By D.S.GIBBS \& I.M.SHAW
}


SINCE publication of the original Scorpio ignition system, about two and a half years ago, many requests have been received for a six volt version for use on cars and motor cycles with six volt electrical systems.

It is, unfortunately, impractical to build a unit to operate on both 6 V and 12 V , so two units have been designed with differing component values, and a number of improvements have been made to reduce radio interference and improve tachometer triggering. Both units give substantially the same performance and can be used on either positive or negative earth cars by altering a few connections.

\section*{PERFORMANCE}

There has been a tremendous rise in interest in capacitor discharge ignition systems over the past few years and their advantages are now widely recognised. The main features are :
(a) Easier starting from cold. The higher spark voltage and faster rise time enables it to fire wet or oiled up plugs. The electronic system can deliver twice the spark voltage of the conventional system into a load of \(470 \mathrm{k} \Omega 2\) and 40 pF -simulating a badly fouled plug.
(b) Smoother running at high speeds. Fig. 6 shows how the electronic system maintains an almost constant spark voltage up to 12,000 r.p.m. for a four cylinder engine ( 8,000 r.p.m. for six cylinder). This promotes better combustion at high speeds, eliminating misfiring and reducing the level of pollutants in the exhaust.
(c) As a result of the improved combustion there is a small improvement in power and fuel economy.
(d) The contact breaker will remain correctly adjusted for up to 20,000 miles. This keeps the car electrically "in tune", with a consequent improvement in fuel economy.
(e) Spark plug wear is reduced. In the conventional system the spark current is unidirectional and causes deposition of metal from one electrode to
the other. In the electronic system the spark current is bidirectional and of much shorter duration. This increases spark plug life and reduces the need for adjustment.

\section*{CIRCUIT DESCRIPTION}

The circuit of both 12 V and 6 V versions is the same and is shown in Fig. 1. Transistor TR1 is not required for positive earth operation but it can be left in circuit if desired. This will enable the unit to be used on either positive or negative earth cars merely by changing a few connections.


Fig. 6. Graph showing the variation of output voltage with engine speed for conventional and capacitor discharge systems


Fig. 1. Circuit diagram of the 12 V version of the Scorpio Mk 2 capacitor discharge ignition system.
Refer to blueprint for full constructional details
Please note: R12 should be 1 W (both versions).
TR3 (6V) should be BFS97 or ZTX550.

The circuit operates as follows (assuming negative earth operation with tag 9 earthed and the contact breaker connected to tag 8)

With the contact breaker closed TRI is saturated and its collector voltage is close to the positive rail. When the contact breaker opens TRI turns off and Cl is rapidly charged by R 3 and diode Dl. This brings TR2 hard on, its collector voltage rises to the positive rail and this positive pulse fires the thyristor CSR1. At the same time TR3 is reverse biased by C3 and remains cut off for about \(0 \cdot 3\) milliseconds. This removes the bias from the inverter and prevents the thyristor from latching on

When the contact breaker points close again TR1 turns hard on, but TR2 remains on for about 0.3 milliseconds-until Cl has discharged through R4 and R5. This effectively prevents contact bounce from retriggering the thyristor at the wrong time. Further noise immunity is provided by C2 and C3, which take about one millisecond to recharge fully after TR2 turns off.

Transistors TR4 and TR5, with the ferrite cored transformer Tl, form a very efficient push pull inverter oscillating at about 2 kHz . The 400 volt a.c. output is rectified by diodes D3 to D6 and charges up the energy storage capacitor C6 in about 1.5 milliseconds. When the thyristor fires this charged capacitor is thrown directly across the primary of the ignition coil. The primary voltage immediately rises to 400 V and this is stepped up to 20 to 30 kV at the secondary, forming an arc between the plug electrodes.

\section*{CONSTRUCTION}

The unit is constructed in a \(7 \frac{1}{4}\) in \(\times 4 \frac{1}{2}\) in \(\times 2\) in diecast box, which acts as a heatsink for TR4, TR5, and the thyristor. Most of the other components are mounted on the printed circuit board shown in Figs. 2 and 3.

Connections to the board are made via turret tags, or the wires can be soldered direct to the board. If turret tags are used they should be expanded with a centre punch after insertion, to lock them in place, and then soldered.

\section*{COMPONENTS}

The diodes, transistors and resistors are all readily available. Capacitors Cl to C 3 are not critical but Mullard type C280 miniature foil types are used in the prototype. The \(1 \mu \mathrm{~F} 440 \mathrm{~V}\) a.c. rated capacitor (C6) is available from many of the advertisers in P.E. The Siemens thyristor is available from A. M. Marshall and Electrovalue who also supply the Siemens pot core. The address of a supplier for the printed circuit board is shown in the components list. The total cost of the unit is around \(£ 8.50\).

\section*{TRANSFORMER CONSTRUCTION}

Details of suitable cores and bobbins are given in the parts list and winding details in Fig. 4. Both types are very similar and give substantially the same perormance.



Fig. 7. Drilling details for the diecast box

\section*{12V version}

First wind on 400 turns of 34 s.w.g. enamelled wire in eight neat layers-taking care not to cross adjacent turns. Insulate the lead out wires with thin sleeving and insulate the whole winding with four turns of thin insulating tape. Then wind on \(12+\) 12 turns of 20 s.w.g. bifilar (i.e. two wires wound together), and lastly \(3+3\) turns of 30 s.w.g. bifilar. Put coloured sleeving on the wires for identification and to protect them from abrasion against the core and insulate the whole coil with several more turns of insulating tape.

It is a good idea to join the two ferrite cups with a very thin smear of Araldite to prevent the core from making an objectionable whine, but test the transformer before doing this as it will be impossible to separate the core afterwards.

\section*{6 V version}

As above, but with a secondary of 300 turns of 34 s.w.g., a primary of \(6+6\) turns of 18 s.w.g. and a feedback winding of \(3+3\) turns of 30 s.w.g.

\section*{ASSEMBLY}

Drill the box as shown in Fig. 7, taking special care to deburr the holes for TR4, TR5, and the thyristor. Mount TR4 and TR5 with mica washers and insulators as shown in Fig. 5 and the transformer T1 with a 2BA bolt and spring washer. The stand-off insulator, and capacitor C7 can now be fixed in place and most of the inverter circuit wired up. Filter LI is fixed with a \(\frac{1}{2}\) in "p-clip" bolted to the case.
Assemble the printed circuit board and mount with 6BA bolts and \(\frac{3}{8}\) in spacers. Then mount the thyristor with a mica washer and a 4BA nylon bolt. It should be possible to solder the gate and cathode leads of the thyristor directly to tags 11 and 4 respectively on the printed circuit board.
It is advisable to fix down the large capacitor C6 either by tying or by gluing as the considerable vibration puts a great strain on the capacitor leads.
Wire up the circuit as shown in Fig. 5 with 14/0076 connecting wire and mount a six-way 5 amp connector block on the outside of the box for the external connections. Fit two plastic covers to the power transistors to prevent accidental short circuits.

Photograph showing the completed Scorpio Mk 2 ignition system

\section*{TESTING}

Before the unit is installed in the car a few simple tests should be performed to ensure that it is operating correctly.

Temporarily disconnect the anode of the thyristor and connect the car battery to terminals \(E\) and \(F\) on the connector block, via a 2 amp fuse. The unit should be heard to "sing" immediately and 400 V d.c. should appear between tags 1 and 4 on the p.c.b. ( 6 volt system-300V).

If the inverter has been wired incorrectly it will probably fail to oscillate at all, but if one base winding has been connected the wrong way round the inverter can oscillate on one transistor only Under these conditions high voltage spikes are produced and up to 800 V may appear across tags 1 and 4.

\section*{INSTALLATION}

\section*{Positive earth}

The wire or wires connected to the SW or "-" terminal of the ignition coil should be connected to terminal \(F\) on the unit instead, and the SW terminal of the coil connected to terminal \(A\). Connect the \(C B\) or " + " terminal of the ignition coil to \(B\), the contact breaker to C , and a good earth from the car body or the positive battery terminal to E .

\section*{Negative earth}

Disconnect the wire or wires to the " + " terminal of the ignition coil and connect them instead to \(E\) on the unit. Connect the " + " terminal of the coil to \(B\) and the "-" terminal of the coil to \(A\). Connect the contact breaker to \(D\) and a good earth to \(F\).

\section*{Cold start coils}

Some cars are fitted with a cold start coil, consisting of a low voltage coil with a series ballast resistor. The ballast resistor is usually connected to one of the terminals of the coil, but may take the form of a resistance cable. It does not make any significant difference to the spark produced by the unit if the resistor is left in series with the coil, but ensure that the supply current for the unit does

not flow through the resistor as there will then be a considerable loss of voltage. In all cases the connection between the starter solenoid and the coil must be removed.

\section*{SEALING}

When the unit is completed and tested and operation is satisfactory the box should be waterproofed to prevent any dampness getting into the electronics. This can be done either using ignition system waterproofing aerosol available under names such as W.D. 40 or "Coldstart", or bath caulking compound. Make sure the grommet holes are sealed as well.

\section*{PRECAUTIONS}

Because of the very high open circuit voltage produced by this unit never attempt to remove the spark plugs whilst the engine is running, since not only is there a danger of breaking down the insulation of the coil, but the experimenter stands a good chance of getting a nasty shock.

In addition the thyristor can be exposed to transient voltages of up to \(1,000 \mathrm{~V}\) under these conditions, and may be destroyed.

\section*{TACHOMETERS}

There are a large variety of tachometers in use, all differing in their exact requirements, so some experimental work may be necessary to obtain the best results.

There are two possible connection points for current operated tachos (the most common type). The lead from the ignition coil to terminal \(A\) on the unit carries a 10 amp short duration pulse and the lead to terminal \(E\) (negative earth) or terminal F (positive earth) carries a wider pulse of 4 amps amplitude. In both cases try coupling the wire through the tacho pickup both ways as the polarity of the pulse may be important.

Regarding voltage operated tachos, the high voltage type should operate if connected to terminal \(B\) on the unit whilst a low voltage type can be left connected to the contact breaker. If in doubt, try both positions.


\section*{NEGATIVE IONS AGAIN}

It seems only recently that we were chatting about the effects that charged atoms and molecules appear to have on us, other animals, and plant life. Since then I have read a number of quite serious reports about these "ionic beasties" and, indeed, discovered in addition that the people at the University of California actually have a faculty there which deals with the subject!

To refresh your memory about ions; generation of airborne jons all begins when sufficient energy is transferred to a gaseous molecule causing it to release an electron. This electron then "climbs aboard" any nearby molecule causing it. in turn, to become charged and in all likelihood convert to a negative ion (by virtue of the charge carried by the electron). The first molecule, however, has, in the process, turned into a positive ion. Umpteen of these molecular collisions occur, with the result that whole layers of species exist having either an overall positive or negative charge.

That, though, is by no means, the end of the story. Water molecules, too, get in on the act and jostle for position around individual ions, so forming somewhat larger ions.

According to Profesor A. Krueger of the University, mentioned earlier, these slightly larger (albeit. still tiny) molecules are of four types, namely \(\mathrm{H}^{+}\left(\mathrm{H}_{2} \mathrm{O}\right)_{12},\left(\mathrm{H}_{3} \mathrm{O}\right)^{+}\) \(\left(\mathrm{H}_{2} \mathrm{O}\right)_{11}, \mathrm{O}_{12}-\left(\mathrm{H}_{22} \mathrm{O}\right)_{n}\), and one with a hydroxyl group attached having the formula \(\mathrm{OH}-\left(\mathrm{H}_{2} \mathrm{O}\right)_{n}\). The \({ }_{n}\) in
the formulae has quite a small value.

Apparently, negative ions are the more mobile, when compared with the positive variety, and, since the Earth itself carries a negative charge, the former are repelled from the surface of the ground.

According to the experts, negative ions are decidedly the "good guys", which bears out our earlier thoughts on the subject. Unfortunately, the positive ("bad guys") are in the majority by something like a factor of \(1 \cdot 2\) to 1 . Pollution, additionally, helps to keep this

ratio looking nasty, because, as you will recall, atmospheric dust has the habit of discharging any negative ions with which it comes in contact.

At the present time we can only produce (and successfully maintain) relatively small quantities of the beneficial ions; i.e., in buildings, or in cubicals where the size of the generator/transmission network can be kept within reasonable bounds. However, there ought

to be sufficient encouragement. for contemplating some really big set-ups. For example; plant growth has been caused to increase by as much as 50 per cent as a result of moderate exposure to air ions. Silkworms, too, hatched out earlier and larval growth was also accelerated. On the other hand, positive ions have been shown to increase the death rate among mice suffering from influenza!

In the final shakedown, it is worthy of mention that negative ions are suspected of lowering the blood serotonin ( 5 -hydroxy tryptamine, 5 HT ) level, and since this is a hormone which, among other things, appears to be connected with the facilitation of neural signals across the synaptic iunctions
between nerves, it is not surpris ing to learn that high negative ion counts can result in profound mental effects. These may range from clearer headedness to a kind of tranquillity.

No amount of persuasion has vet encouraged my editor to change his mind about putting a taboo on a special high-voltage negative ion generator 1 intended to include. It could no doubt, be lethal in the wrong hands, so I guess his decision is the right one!

\section*{MAGNETO-CHEMISTRY}

Still on the subject of chemy's, I notice that R. Skorsky of the Alabama University has discovered an interesting effect in ferric oxide, \(\mathrm{Fe}_{2} \mathrm{O}_{3}\). Haematite, for that's it's other name, can, according to him, be reduced at much lower temperatures than previously achieved by simply surrounding the sample in an a.c. field of a few hundred gauss.

Skorsky has a theory that the cause of this effect is due to the increased pressure of hydrogen at the surface of the material resulting from its attraction to the magnetised iron (due to initial reduction of some of the ferric oxide at the beginning of the reaction).

Interestingly enough, C. Peters. over here, at the University of Surrey seems to think that the effect can be expected. His proposal amounts to the fact that since there is a considerable difference in the magnetic properties of the non-reduced material and final product, and because the reaction is going in a "direction" which favours a ferromagnetic substance as an end result, anyway, the outcome will be an increased tendency towards production of a magnetic material when a field of sufficient intensity is available. Sounds reasonable!

\section*{TAILPIECE}

Washington, you'd think, had received its share of trouble. But it would seem the wags haven't finjshed with 'em yet. Only yesterdav I came across this odd bit of logic.


\title{
A.C. DFTECTOR PROBE
}

\section*{By B. BOWLES}

II is often necessary, particularly in radio circuits, to check for the presence of high frequency oscillations but most amateurs do not possess test equipment capable of responding to frequencies greater than a few megahertz. The probe described here will respond to frequencies between 200 Hz and 30 MHz , allowing use in general a.c. signal tracing. Any a.c. of a few millivolts or more will light a small light emitting diode (l.e.d.). The whole instrument can be held in one hand.

\section*{CIRCUIT DESCRIPTION}

The complete circuit is shown in Fig. 1. Transistors TR1 and TR2 form a high gain a.c. amplifier which also rectifies the signal to drive the l.e.d. (D2). The configuration is known as a d.c. feedback pair. TRI is biased just into conduction by feedback from the emitter of TR2. The preset potentiometer VR1 is used to set the d.c. conditions for the circuit. Components R5 and C2 and C3 form a filter which reduces the a.c. feedback so that the a.c. gain of the circuit is not reduced.
The input impedance is roughly equal to the value of R2 since C2 and C3 form a low impedance path for a.c. C3 is used to compensate for the inductive effect of C2 at high frequency. Diode D1 protects the input from negative voltage swings.

\section*{RECTIFICATION}

The diodes in the emitter of TR2 serve to rectify the signal in the following manner. When the input swings positive TR1 will conduct tending to lower the voltage at TR2 base and by emitter follower action at TR2 emitter as well. The diodes cease to conduct and the voltage at TR2 emitter drops causing the current fed back to TR1 to drop, counteracting the effect of the positive input. If, however, the input goes negative, TR1 will conduct less causing TR2 base to rise. The diodes will, however, maintain the voltage at TR2 emitter almost constant so there will be no negative feedback to the base of TR1
The circuit thus has high gain when the input goes negative but low gain when it goes positive.

Thus an a.c. input will cause TR2 to conduct and hence pass more current through the l.e.d. The resistor in parallel with the l.e.d. allows enough current to pass through the transistor to keep the diodes conducting and the resistor in series limits the maximum l.e.d. current to less than 10 mA .
The circuit is so stable that the battery voltage can fall to 7 V without affecting performance.

\section*{COMPONENTS . .}
\begin{tabular}{llll} 
R1 & \(4.7 \mathrm{k} \Omega\) & R4 & \(1 \cdot 2 \mathrm{k} \Omega\) \\
R2 & \(4.7 \mathrm{k} \Omega\) & R5 & \(4.7 \mathrm{k} \Omega\) \\
R3 & \(120 \Omega\) & & \\
All \(\pm 5 \% \frac{1}{3} \mathrm{~W}\) carbon & &
\end{tabular}

\section*{Potentiometer}

VR1 \(5 \mathrm{k} \Omega\) linear sub-miniature skeleton horizonta mounting

\section*{Capacitors}

C1 \(0.22 \mu \mathrm{~F}\)
C2 \(10 \mu \mathrm{~F} 6.3 \mathrm{~V}\) tantalum bead
C3 \(1,000 \mathrm{pF}\) ceramic
C4 \(47 \mu \mathrm{~F} 10 \mathrm{~V}\) elect.

\section*{Diodes}
D1
1N4148
D2
TIL209 light emitting diode
D3
OA91
D4

\section*{Transistors}

TR1, TR2 BC107 or BFY90 (2 off)

\section*{Miscellaneous}

S1 Push-to-make, release-to-break push-button B1 9V PP3 battery
Veroboard 0.1 in matrix \(2 \mathrm{in} \times 0.9 \mathrm{in}\)
Veropins, 6BA and 8BA nuts and bolts and washers, wire, suitable case (see text)


Fig. 1. Circuit diagram of the A.C. Detector Probe


Brass cheese head screw
SENSOR 68 A \(\times 20 \mathrm{~mm}\) Iong


Fig. 2. Layout of the components on the Veroboard and interwiring details


Photograph showing the completed A.C. Detector Probe

\section*{CONSTRUCTION}

The circuit for the A.C. Detector Probe can be made very compact so that the whole unit can be held in the hand, making it convenient for testing equipment. All the components except the capacitor Cl are mounted on a small piece of Veroboard as shown in Fig. 2. Veropins are used for wire connections to the board. The l.e.d. is mounted on two pins higher than the rest of the board so that it shows through the side of the probe case. Take care not to overheat the l.e.d. when soldering.

Only one end of Cl is connected to the board.

\section*{SETTING UP}

Before mounting in the case VR1 must be set up.
Connect the battery to the board. The l.e.d. should light with VRI at one end of the track and extinguish at the other end. If the l.e.d. dims but does not go out, TRI has low gain and should be replaced.

VRI should be adjusted so that with no input D2 is just faintly lit. Any a.c. input with a frequency

\section*{Photograph showing the construction of the A.C. Detector Probe before assembly}

of 200 Hz to 10 MHz should now cause the l.e.d. to light.

The l.e.d. should flash each time the power is applied, this being due to the charging of \(C 2\).

\section*{CASE CONSTRUCTION}

The prototype case was made from an empty butane lighter fuel container of medium size but any aluminium tube will do. Great care should be taken that there is no gas left in the container.

First pierce the container about five inches from the tip or nozzle. This should be done well away from naked lights, preferably outdoors. Then cut the case into three parts as shown in Fig. 2.

Do not drill the holes yet. After the cutting and any necessary smoothing of the edges, the body of the probe, the end cap and a 19 mm ring remain. This ring should be cut with scissors, its ends overlapped and then pushed into the probe body. Mark the overlap, remove the ring and cut off the overlap so that when re-inserted into the tube it fits with the ends butted together. The ring can now be glued into the body leaving about 6 mm protruding from the body.

\section*{END CAP}

This now makes a push-fit seating for the end cap, and provided the cuts have been made square gives the probe a good appearance.

The end cap should now be pushed on and held with adhesive tape and the holes drilled.

The holes in the end cap should now be enlarged into slots so that the end cap can slide over the two bolts which secure it. A hole is also drilled in the end cap to take the push button switch.

The plastic insert in the front of the tube should alse be enlarged to 3 mm .

\section*{FINAL ASSEMBLY}

The free end of Cl is soldered to the head of a 12 mm 6 BA bolt. The solder tag on the mounting bolt should be soldered to the negative line on the board. The battery is then connected up. Some insulating tape is stuck inside the tube in case the circuit should touch, and the board pushed in making sure that the bolt on Cl goes through the end of the tube. The battery is then pushed in with a piece of card between it and the board to stop the metal case shorting the board.

The end cap is now fitted and nuts fitted to the bolts holding the end cap and the probe bolt. Attachment of the earth lead and crocodile clip completes the unit.

\section*{USE AND APPLICATIONS}

The A.C. Detector Probe is very sensitive and its frequency response extends up to several megahertz. lf high voltages are to be measured a resistor can be connected to the probe bolt to act as an attenuator.

Bandwidth can be extended by eliminating D1 and replacing TR1 and TR2 with BFY90 transistors which give high sensitivity up to about 70 MHz . Above this frequency the l.e.d. may actually dim when connected to a.c. this being due to phase shift causing the normal rectifying action to be reversed. This is just as valid an indication of the presence of a.c.

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 \\ \title{
CRYSTAL \\ \title{
CRYSTAL byda peratue
}

THE PURPOSE of a crystal calibrator is to provide signals at standard frequencies over the medium and short wavebands, in order to allow accurate calibration of a radio receiver dial.

Providing the output from the calibrator is nearly a square wave it will be rich in harmonics (i.e. multiples of the fundamental frequency). A good calibration oscillator of fundamental frequency 200 kHz will have harmonics which are reasonably strong right up to the upper limit of the short wave band ( 30 MHz ).

Using a crystal means that the fundamental frequency is predetermined, and extremely stable with respect to time and temperature.

In order to allow the signal to be easily identified on a receiver an audio oscillator has been included to modulate the r.f. signal.

\section*{CIRCUIT DESCRIPTION}

The full circuit diagram is shown in Fig. 1 and can be seen to be divided into two sections, the audio oscillator and the crystal controlled r.f. oscillator. The active elements \({ }^{\bullet}\) used in the unit are contained in two integrated circuits type \(\mu\) L914 whose internal circuits are shown in Fig. 2.

The \(\mu \mathrm{L} 914\) is intended for use in logic systems and is one of the family of Resistor Transistor Logic (RTL) circuits. It can be adapted for use as an amplifier by inserting a biasing resistor between the collector and the existing base resistor of one of each pair of transistors and ignoring the other, thus forming two, one-transistor amplifiers from each i.c.

\section*{CRYSTAL OSCILLATOR}

A crystal has the property that at one particular frequency (known as the series resonant frequency) its impedance (resistance to current flow) becomes very low while at all other frequencies its impedance is high.

In the oscillator circuit used in the calibrator the two, one-transistor amplifiers in the \(\mu \mathrm{L} 914\) are connected in series via the crystal and the capacitor C6.

At the series resonant frequency the impedance of the crystal is sufficiently low for the positive feedback between the two stages to maintain oscillation.

Because of the large positive feedback the switching between the two states (transistor on and transistor off) is rapid so that the output is almost square producing the large number of harmonics required. The 200 kHz crystal used in the prototype gave enough harmonics to cover the long and medium bands.

\section*{AUDIO OSCILLATOR}

The other i.c. and its associated components form an audio frequency oscillator with a frequency of about 2 kHz . When applied to the crystal oscillator

COMPONENTS . . .
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|l|}{Resistors} \\
\hline R1 \(2.2 \mathrm{k} \Omega\) & \\
\hline R2 \(47 \mathrm{k} \Omega\) & \\
\hline R3 \(2.2 \mathrm{k} \Omega\) & \\
\hline R4 \(100 \mathrm{k} \Omega\) & \\
\hline R5 \(100 \mathrm{k} \Omega\) & \\
\hline R6 \(33 \mathrm{k} \Omega\) & \\
\hline R7 \(470 \Omega\) & \\
\hline R8 \(100 \mathrm{k} \Omega\) & \\
\hline \multicolumn{2}{|l|}{All \(\pm 10 \% \pm\) watt} \\
\hline \multicolumn{2}{|l|}{Capacitors} \\
\hline C1 \(0.1 \mu \mathrm{~F}\}\) ceramic & \\
\hline C2 \(0.1 \mu \mathrm{~F}\}\) ceramic & \\
\hline C3 120pF & \\
\hline C4 \(0.047 \mu \mathrm{~F}\) ceramic & \\
\hline C5 \(100 \mu \mathrm{~F} 10 \mathrm{~V}\) elect. & \\
\hline C6 2,700pF & \\
\hline C7 \(0.01 \mu \mathrm{~F}\) & \\
\hline \multicolumn{2}{|l|}{Integrated Circuits} \\
\hline IC1, IC2 \(\mu\) L914 or equivalent (2 off) & \\
\hline \begin{tabular}{l}
Crystal \\
XL1 200 kHz crystal with holder (Henry' type FT241A.)
\end{tabular} & Radio \\
\hline Miscellaneous & \\
\hline S1 4-pole 3-way rotary & \\
\hline B1 9V battery type PP4 with clips & \\
\hline SK1 Coaxial panel mounting socket & \\
\hline 2.8 in \(\times 1.4\) in Veroboard 0.1 in matrix & \\
\hline Aluminium case approx. \(4 \mathrm{in} \times 3\) 交 \(\mathrm{in} \times 2 \mathrm{in}\) & \\
\hline
\end{tabular}


Fig. 2. Circuit diagram showing the internal circuit of the \(\mu \mathrm{L914}\) integrated circuit


Fig. 3. Constructional details of the Crystal Calibration Oscillator

via R6 it varies the amplitude of the r.f. signal so that an audio signal of 2 kHz will be heard when the receiver is tuned correctly.

The two amplifiers in the i.c. are connected in series to form an amplifier whose output is in phase with its input. The components C1, R1 and C2, R3 form what is known as a Wien Bridge, and it is these components that determine the frequency.
The switch therefore has three positions, OFF, OSC which gives the unmodulated crystal frequency and MOD which gives this frequency modulated by the oscillator.

Resistor R7 and capacitor C5 are included to reduce the supply voltage down to the 5 V required by the i.c.s.

\section*{CONSTRUCTION}

Construction is simple, all the components except the crystal, switch and output socket being mounted on a piece of Veroboard as shown in Fig. 3.

Any small aluminium case will do providing it is of sufficient size to accommodate all the components. It is wise to obtain a socket with the crystal as soldering to it directly could cause damage. A small bracket to hold the crystal can be mounted on the side of the box as can be seen in the photograph.
The switch used in the prototype is a four-pole three-way type with only two poles used.

\section*{USING THE CALIBRATOR}

Once the unit has been completed, and all wiring checked for mistakes, a length of lead should be connected to the output socket and the unit turned on.
If the lead is now placed near a medium wave radio receiver, and the receiver is tuned over the band, strong signals from the calibrator should be received at intervals across the dial. Switching on the modulation should produce an audible note on these signals.
Using a 200 kHz crystal, harmonics will be generated at \(600 \mathrm{kHz} .800 \mathrm{kHz}, 1 \mathrm{MHz}, 1.2 \mathrm{MHz}\), and 1.6 MHz . To find out which is which the receiver should be tuned to a known frequency (say Radio 1 on 1.214 MHz ) then the dial turned until the nearest signal from the calibrator is received '(in the case above this will obviously then be 1.2 MHz ). Other frequencies can then be ascertained by counting the number of harmonics above or below the known frequency, as each corresponds to a difference of 200 kHz .

This method of identification can be used on arry band where the exact frequency of a station is known.


A selection of readers' suggested cincuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.
This is YOUR page and any idea published will be awarded payment according to its merits.

\section*{OPTICAL COMMUNICATION SYSTEM}

The circuit described here is for a novel communication system using light beams to carry audio signals. Light emitting diodes (1.e.d.s) are particularly suitable for producing modulated light beams as they produce an intense concentrated spot


Fig. 1. Transmitter circuit diagram

\section*{BOILER GAS VALVE OPERATION AT LOW D.C.}

PERHAPS the enclosed circuit idea would be of interest to your readers in view of impeding power cuts this winter. It allows a gas boiler mains operated gas valve, to be operated from a low voltage d.c. supply. The circuit has low battery consumption and low break current to minimise thermostat contact erosion.

I hesitate to recommend exact component values as the circuit was developed hurriedly to be ready for power cuts. In addition it was intended to run from a car battery so I did not bother to optimise the oscillator circuit or the capacitor and charge resistor values for minimum consumption.

I used an available Ferroxcube in the oscillator circuit and I do not know the exact turns ratios. I do not know how much variation exists between individual gas valves so that some experimentation by readers will therefore be required in both these areas.

Fig. 1 shows the circuit where R 1 and C 1 are included to prevent possible, radio interference.

In operation, Sl is first closed charging C5 to about 120 V . After a minute or so S 2 is operated. The charge on C5 operates the gas valve, the voltage of which is prevented by D1 from falling much
of light, have a linear current/light output relationship, and have a good high frequency response (the type of l.e.d. with a plastic diffusing dome is not suitable for use in this circuit).

The circuit shown in Fig. 1 will modulate the light from the l.e.d. using a cystal microphone or a loudspeaker output. IC1 acts as an a.c. amplifier with an input impedance of about 1 M !. R 1 and R 2 are biasing resistors and together with IC1 and TR1 they produce a d.c. of 30 mA in the l.e.d. VR1 and C2 control the a.c. feedback causing the input signal to be amplified and superimposed upon the bias.

To obtain the maximum range from the transmitter the optical system must be efficient. Either a convex lens or a concave mirror as shown in Fig. 2 can be used to convert the l.e.d. output into a parallel beam.

For efficient reception the light collecting area must be large; the author used a large concave makeup mirror costing 40 p , but better quality mirrors will give better results.

The received light is concentrated onto a sensitive photo-Darlington transistor as shown in Fig. 3.

Because the transistor is so sensitive the load resistor must be low to prevent saturation in daylight conditions.

At short range the signal across the load resistor is adequate to drive a crystal earpiece, though for longer range an a.c. amplifier and a loudspeaker are needed. The phototransistor output impedance is \(1 \mathrm{k} \Omega\) and most commercially available audio amplifiers are suitable.

The overall distortion is quite low, speech being excellent and music quite acceptable. The main cause of interference is mains lighting but this can be minimised by careful design of the optical system.

Transmitter and receiver must be accurately aligned and the transmitter gain should be carefully adjusted by means of VR1 to give maximum modulation without clipping. Clipping causes obvious distortion and is easily recognised.
J. N. Jones, Cheshire.


Fig. 2. Typical optical system for the transmitter


Fig. 3. Receiver circuit diagram
below the battery voltage. When the thermostat cuts out C5 recharges ready for the next operation.

C4 was included to adjust the charge rate of C5; it obviously affects the oscillator also and should not be needed in a fully developed design. D1 and D2 should have a minimum p.i.v. of say 200 V and 400 V respectively. D3 and D4 can be any silicon diodes.

I found that my Potterton boiler gas valve included a rectifier and that it closed at about 70 V d.c. and held in at voltages below 11 V . The consumption was about 2 mA at 12 V d.c.

For better economy of operation, a switching circuit might perhaps be devised to switch off the oscillator when the gas valve is energised but I have not tried this.
C. J. Collins,

Fig. 1. Circuit for operating gas valve from low voltage d.c. supply



Fig. 1. Circuit diagram of the slow timebase oscillator

THE sawtooth generator shown in the circuit diagram Fig. 1 was designed to extend the timebase range of a general purpose oscilloscope down to 1 Hz . As well as the linear sawtooth with fast flyback, there is provision for an external synchronising signal and flyback blanking output.

The timing capacitor Cl of the relaxation oscillator, TR2, is charged by a constant current source from TR1. The base potential of TR1, controlling the constant current, is derived from the divider network R1, VR1 and R2. The Zener diode across the divider keeps the voltage at a constant level.

Since the sawtooth output at the emitter of the unijunction is at high impedance, the output is fed to an emitter follower to buffer the output against possible low impedance loads. The output should, however, not be taken to loads of less than \(20 \mathrm{k} \Omega\) ? impedance.

The synchronising input has a constant input resistance of \(2 \mathrm{M} \Omega\) and a portion of the signal is taken to an emitter follower via VR2 and C2. The effect of the emitter follower output is to modulate the supply voltage to the unijunction in step with the sync signal so that the sawtooth tends to be locked in frequency to the sync signal or a multiple thereof.

For adequate synchronisation even at the lowest frequencies, the input impedance of the emitter follower must be greater than the reactance of the capacitor at the lowest frequency.

The voltage rating of C 2 should be sufficient to block the highest d.c. voltages likely to be encountered.

Adequate sync is available with a 0.3 to 1 V r.m.s. input, but if higher voltages are used a diode should be wired as shown dotted, see Fig. 1, to prevent transistor emitter/base breakdown.

The frequency range of the oscillator is 1 to 60 Hz , but could be extended up to 20 kHz by varying the value of the timing capacitor C1. Capacitor C1 should be a low leakage type to ensure good linearity at low frequencies.

One of the many applications would be to use the oscillator for monitoring heart pulses in conjunction with the "Biological Amplifier" published in the January and February 73 issues (no back issues available). Care should be taken to ensure that the output from the Bio-amplifier is a.c. coupled to the vertical input of the monitoring 'scope.
M. H. George,

Upminster. Essex.

\section*{AN IMPROVISED TAPE GUIDE}

T"HE drawing in Fig. I shows an improvised tape guide with height adjustment.
The guide itself is a \(\frac{1}{4}\) in brass binding post and is of the type which is threaded right through. These are obtainable from most well-stocked stationers.

The guide is simply fixed and adjusted by a BA bolt and three nuts.

Paul A. Koh, Sunbury-on-Thames.

Fig.1. Construction of the improvised tape guide


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\section*{PRACTICAL}


APRIL ISSUE ON SALE MARCH 8, 1974

\title{
Boolean Alydinl-1
}

\author{
By B. J. Wood
}

\section*{A two part article providing an introduction to switching theory and concluding with a practical design of a simple adder/subtractor}

George Boole published a treatise "An Investigation of the Laws of Thought" discussing the system now called Boolean Algebra over a century ago in 1861. The basis of his concept is that a statement in words may be represented in the first place by a letter of the alphabet.

Thus we can say that the letter A represents the statement "It is Sunday", and the letter B "It is sunny." These are positive statements and Boole assigned them a value 1 .

Using his system it is very simple to identify the opposite statements "It is NOT Sunday" and "It is NOT sunny" by simply putting a bar over the chosen letter thus \(\overline{\mathrm{A}}\) or \(\overline{\mathrm{B}}\). This time Boole assigns the value 0 to these negative statements.
To test the system let us assume that we ask the question "Can I mow the lawn?", having set the conditions that one can, but only if it is sunny and also a Sunday.
Thus for a sunny Sunday we have both A and B , or shown mathematically, A.B, where both A and \(B\) are 1. Clearly the answer is I, or positive. If either of the conditions is a negative then the answer will be negative. Thus \(\overline{\mathrm{A}} \cdot \mathrm{B}=0\) and \(\mathrm{A} \cdot \overline{\mathrm{B}}=0\) and it is not suitable to mow the lawn.
This type of decision making is called an AND function in which the factors have been multiplied together.
The condition could be "It is Sunday" or "It is sunny" in which case an addition of the two values would be required. If either had a value of I the result would be I. This is called an OR function and in the case of the OR when both are 1 the rather odd result of \(1+1=1\) arises but here the AND and OR are interchangeable.


If the condition "Sunday and dry" is just as suitable for mowing the lawn as "Sunday and sunny" the decision to mow the lawn or not may be stated as A.B + A.C where C means "It is dry". Boolean algebra may be manipulated like any other algebra so that \(\mathrm{A} \cdot(\mathrm{B}+\mathrm{C})\) means the same as \(\mathrm{A} \cdot \mathrm{B}+\mathrm{A} \cdot \mathrm{C}\), giving the same arithmetic result.

If there are many conditions and alternatives the algebra can become quite involved. Generally, it is enough to work out one or two decisions at a time and build up the logic gradually.

\section*{LOGIC DIAGRAMS}

In logic of the electronic kind, diagrams are useful to describe the flow of information. In the initial stages such diagrams contain only simple elements whilst later a whole group of elements may be put into one box and given a label in the same way that a collection of components may be labelled "amplifier". Fig. 1 shows the two ways of connecting AND and OR boxes which will give the same decision, to mow or not to mow the lawn.
The symbols used are fairly simple, being in the form of "D" shapes with inputs to the left and output to the right, a figure in the symbol denotes the number of inputs which must assume a logic 1 state for the output to be logic 1. An \& sign in the symbol denotes an AND element.

Thus the operation to decide to mow the lawn under the Sunday and sunny or Sunday and dry conditions can be determined from these diagrams. Notice that the simplification of A.B + A.C to \(\mathrm{A} .(\mathrm{B}+\mathrm{C})\) also simplifies the diagram.

\section*{LOGIC ELEMENTS}

Logic integrated circuits are available in many forms, simple and complex. They are very convenient packages, reducing external wiring to a minimum. Their low cost allows the constructor to build equipment which, a few years ago, would have been too expensive in terms of time and money. For the moment, however, elements built from discrete components will be discussed and related to the i.c.s as convenient. It does not matter that these devices are not identical as long as they react in the same way.

\section*{AND, OR, NOT ELEMENTS}

These are the basic devices. More complex devices may be built from them. The AND may be made from diodes and a resistor as in Fig. 2a.

If either of the switches is closed representing an input of a 0 function then the whole of the voltage available from the supply is dropped across R1 and the output is at earth potential. Obviously, closing both switches has the same effect. It is only when both switches are open that a voltage is available at the output terminal.

The presence of this voltage is called logic 1. In its absence the state is logic 0 . Why "logic"? Normally, the switches in the diagram are replaced by the outputs of other elements.

In practice absolute voltage levels are not possible at reasonable cost and so the design must allow for fluctuation. Thus logic 0 may be 0.5 V or less and logic 1 anything over 2 V .

To following elements these values are taken to represent 0 or 1 . There is an electrical no-man'sland in between the two values which is avoided by the design and proper use of the devices involved.

\section*{AND}

To return to the AND element, there is a table at 2 a in which the states of A and B are given together with the resultant output for the four possible states of input. Called a truth table, this device shows that for a 1 output all inputs must be at logic 1. For any other combination the output is at logic 0 . This applies for two, three or twenty inputs. Note that, to an AND input, an open circuit is a logic 1.

\section*{NOT}

The NOT is shown at Fig. 2b It is simply an inverter in which, if a logic 1 voltage is applied at the base of the transistor it is switched hard on, the collector bottoms and the output goes to logic 0 . If the base is at logic 0 the output is up near the rail voltage.

The second diagram at Fig. 2 b is much more like an i.c. inverter in that the transistor is held on with a logic 0 output, unless presented with a logic 0 input. Like the AND, an open circuit input is treated as logic 1.

\section*{SYMBOLS}

There are conventional symbols used to denote the various elements we have discussed so far as well as others to follow. They are used to avoid the need for either discussing or indeed even showing the actual components used in a particular element such as the diodes or transistors already noted.

Thus the elements AND and NOT, shown in detail at Figs. 2a and 2b may be shown as the two separate ' \(D\) ' shapes joined together in Fig. 2c. Here the AND element is that with the ampersand (\&) inscribed. This is intended to show that the element has inputs which have to be at logic 1 for the output to go to logic 1.
The second element with the small circle on the output line is the NOT element of Fig. 2c. Here the small circle denotes the logic 0 output obtained if the input goes to a logic 1

\section*{NAND}

The two elements at Fig. 2c denote NOT AND which is normally known as NAND. In fact this has a new symbol of its own as shown to the right in Fig. 2c, the small circle denoting the NOT output and the ampersand the requirement for logic 1 inputs for this. The truth table shows the outputs obtained for various inputs and it can be seen that only when all inputs are logic 1 will the output be logic 0 . Of course, as before the logic 1 input includes an open circuit.


In its simplest form the OR element is made from diodes as shown at Fig. 2d. As can be seen these point in the opposite direction from those in the AND element. If a logic 1 appears at any input it becomes available at the output. The OR treats an open circuit as logic 0 . In diode form the resistance of the load must be small enough to look like a logic 0 to the following element but large enough to demand only a reasonable current from the input source. The second diagram of Fig. 2d overcomes this difficulty by the introduction of two NOTs.

In this case the first element is a basic threshold element and the number inside denotes the number of inputs which must be at logic 1 for the output to be at logic 1. This element is followed by the two NOT elements and the whole may be shown as below.

\section*{NOR}

A NOR element is an OR followed by one inverter element as shown at Fig. 2e. Only when all the inputs are at logic 0 is the output at logic 1. Under any other condition, as can be seen from the truth table the output is at logic 0 .

\section*{INVERTED INPUTS AND OUTPUTS}

Suppose that an AND is used with an inverter in each input so that 0 becomes 1 and vice versa. Fig. 2f truth table shows that it becomes a/NOR. If, additionally, the output is inverted it then becomes OR. The reader should treat all the other truth tables in à similar fashion. It will be seen that any element may be made to exhibit the characteristics of the other three.

For what follows in this text the important point is that a NAND with inverted inputs becomes an OR. In general the devices to be used will be NANDs with diodes as ANDs when convenient plus NOTs. In the beginning it will be, perhaps, easier to recognise that an OR is required and then use a NAND with inverters in the inputs.

\section*{BINARY NUMBERS AND ARITHMETIC}

In the kind of logic under discussion there are only the two states 1 and 0 . Any process to be handled by the logic must therefore be described in these terms. In everyday arithmetic the radix of 10 is used with ten possible digits of 0 to 9 . But if, as with Boolean logic, 2 is used as the radix, only the digits 0 and 1 are possible.

In the scale of 10 the number 286 can be written \(\left(2 \times 10^{2}\right)+\left(8 \times 10^{1}\right)+\left(6 \times 10^{\circ}\right)=200+80\) \(+6=286\).

In the scale of 2 the number 1101 means \(\left(1 \times 2^{3}\right)\) \(+\left(1 \times 2^{2}\right)+\left(0 \times 2^{1}\right)+\left(1 \times 2^{\circ}\right)=8+4\) \(+0+1\). (A number to the power of 1 is itself. Any number to the power of 0 is 1 ).

As a point of interest, decimal-to-binary conversion can be achieved as follows.

\section*{CONVERSION}

Convert 83 to binary. This may be done mechanically by dividing the number 83 continuously by 2 ignoring remainders until the answer of 1 is reached.

Write (-) against odd answers and (0) against even answers. The result looks like this.
\begin{tabular}{ll}
83 & - \\
41 & - \\
20 & 0 \\
10 & 0 \\
5 & - \\
2 & 0 \\
1 & -
\end{tabular}

Il now the paper is turned so that the (-) against the final answer of 1 is at the left then this is the result in binary:-
\[

\]
where \(1010011=64+16+2+1=83\).
Conversely, to convert binary notation to its decimal counterpart simply multiply the left-most digit by two and add the following binary digit. Multiply the result by two adding the next binary digit and so on until the end is reached as below.
\[
\begin{array}{rrrrrr}
1 & 1 & 0 & 1 & 0 & 1
\end{array}
\]

The final figure, in this case 53 , is the correct answer.

\section*{ADDING BINARY NUMBERS}

The rules for adition of two binary digits (bits) are \(0+0=0 ; 0+1=1 ; 1+1=0\) and carry 1. Where there is a carry from the previous position it requires the following rules, \(0+0+1=1\); \(0+1+1=0\) and carry \(1 ; 1+1+1 \doteq 1\) and carry 1.
Using the rules to add 6 to 7 -
Binary Decimal
\[
\begin{aligned}
& 111=7 \\
& 110=6 \\
& \hline 1101=13 \\
& \hline
\end{aligned}
\]

\section*{SUBTRACTION}

The rules for subtraction of two binary digits (bits) are \(0-0=0 ; 1-1=0 ; 0-1=1\) and borrow \(1,1-0=1\). Where there is a borrow the following applies, \(0-0-0=0\); \(1-0-0=1\); \(0-1-0=1\) and borrow \(1 ; 1-1-0=0\); \(1-1-1=1\) and borrow 1 , etc.

Thus if we subtract 3 from 5:
Binary Decimal
\[
\begin{aligned}
& 101=5 \\
& 011=3 \\
& \hline 010=2 \\
& \hline
\end{aligned}
\]

Of course, problems arise when larger numbers have to be subtracted from smaller numbers. Using decimal notation, the human animal in fact reverses the process, subtracts the smaller from the larger and then calls the result a "negative."

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\hline + & 10\% & 3.3Mn-10Ms & E12 & Ip & 0.8p \\
\hline \(\frac{1}{2}\) & 2\% & 10n-1M & E24 & \(3.5 p\) & \({ }^{3 p}\) \\
\hline \(\frac{1}{2}\) & 10\% & \(18-3.9 n\) & E12 & 1 p & 0.8p \\
\hline 4 & 5\% & 4.7日-1Mn & E12 & 1 p & 0.8p \\
\hline 4 & 10\% & \(1 \Omega-10 \Omega\) & E12 & \(6 p\) & 5.5 p \\
\hline
\end{tabular}

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\(470 / 40,680 / 25,1,000 / 16,1,500 / 10,2,200 / 6 \cdot 3,18 \mathrm{p}, 330 / 63,680 / 40,1,000 / 25,1,200 / 16\), \(2,200 / 10,3,300 / 6 \cdot 3,4.700 / 4,21 \mathrm{p}\).

\(\begin{array}{rl}0.1 \mu \mathrm{~F} & 35 \mathrm{~V} \\ 0.22 \mu \mathrm{~F} & 35 \mathrm{~V}\end{array} 2.2 \mu \mathrm{~F} 35 \mathrm{~V}\)
\(0.22 \mu \mathrm{~F} 35 \mathrm{~V}\)
\(\begin{array}{cc}0.47 \mu \mathrm{~F} & 35 \mathrm{~V} \\ 1.0 \mu \mathrm{~F} & 35 \mathrm{~V}\end{array}\)
\(\begin{array}{ll}4.8 \mu \mathrm{~F} & 25 \mathrm{~V} \\ 10 \mu \mathrm{~F} & 25 \mathrm{~V}\end{array}\)
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\hline
\end{tabular}

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Whilst a machine can be made to do this it is not part of the present discussion as it is complex and requires very lengthy explanation.

Take the case of subtracting 7 from. 3. In decimal the answer is obvious, it is minus 4 ( -4 ). Equally, if you take 5 from 2 the answer is -3 . In binary terms using the methods so far discussed this is not so. Thus in the case of \(3-7\) :
\begin{tabular}{l}
011 \\
\(\frac{111}{\text { (1) } 100}\) \\
\(\frac{\text { And }}{}\) \\
010 \\
101 \\
(1) 101
\end{tabular}

In both cases there is a borrow 1 with nothing to subtract it from and in fact even if there were more positions to the left the resultant row would be filled with figure 1's with an unsatisfied borrow 1 emerging at the far left.

The same thing happens in decimal notation if one persisis in trying to subtract using this type of approach. Thus in decimal the results could be shown as (1)6 and (1)7, the figure in the bracket being the unsatisfied borrow.

In fact the two answers 6 and 7 are the tens complements of the correct answers, subtract the 6 and the 7 from 10 and you have, in positive form, the correct answer. .

Similarly, with the binary computation the results must be subtracted from 8 which, in binary is 1000 There are two snags to this situation. First the operator needs to know before the event that a negative result will arise and re-arrange the inputs.

\section*{COMPLEMENT}

The word complement was mentioned. This is the key to the answer. In the 1 and 0 binary system the 1 s complement of 1 is 0 and vice versa.

Ignoring the unsatisfied borrow in the two binary answers, if 1 is changed to 0 and vice versa 100 becomes 011 and 101 becomes 010 or 3 and 2 in decimal, both wrong by being 1 too few. What is


IMPROVING YOUR HI-FI

\section*{By John Earl}

Published by Fountain Press
203 pages, \(8 \underset{4}{4} \mathrm{in} \times 5 \frac{1}{2} \mathrm{in}\). Price \(£ 3.00\)

THE title of this book is slightly misleading as the theme is not really improving hi-fi systems but getting the best out of them. Written in a nontechnical style, it appears to be directed towards the people who are not satisfied with low-fi but are not \(a u\) fait with all the technicalities and jargon of high-fi and are perhaps not getting all they could from their set-up.

This is strictly a practical book dealing with things like the importance of correct earthing, the need to
really needed is the complement of 7 which can be arranged by subtracting 1 before the complement is taken. To do this the unsatisfied borrow from the left is carried round to the right to form part of the subtraction. Using the examples again-
\begin{tabular}{|c|c|c|}
\hline 011 & 010 & 迷 \\
\hline 111 & 101 & Subtract \\
\hline (1) 100 & (1) 101 & \\
\hline \(\rightarrow 1\) & \(\rightarrow 1\) & Subtract \\
\hline 011 & 100 & Answers \\
\hline 100 & 011 & Complements, both in fact \\
\hline
\end{tabular} negative.
The complements give the answer in positive form but it is known that the answer is negative because the mode of operation is subtract and there is the borrow 1.
The following table gives all the possible results for the system. The reader is advised to work out columns \(b, c\) and \(d\) to gain an understanding.

\begin{tabular}{lllll}
0 & 0000 & & & \\
1 & 0001 & 1111 & 1110 & 0001 \\
2 & 0010 & 1110 & 1101 & 0010 \\
3 & 0011 & 1101 & 1100 & 0011 \\
4 & 0100 & 1100 & 1011 & 0100 \\
5 & 0101 & 1011 & 1010 & 0101 \\
6 & 0110 & 1010 & 1001 & 0110 \\
7 & 0111 & 1001 & 1000 & 0111 \\
8 & 1000 & & & \\
9 & 1001 & & & \\
10 & 1010 & & & \\
11 & 1011 & & & \\
12 & 100 & & & \\
13 & 1101 & & & \\
14 & 1110 & & &
\end{tabular}

\section*{Next month : Constructional details of a simple adder/ subtractor will be given}
maintain correct dynamic ranges throughout the system, and the meaning of specifications.

For the non-technical enthusiast this book would be essential reading before he considers buying expensive hi-fi equipment
S.R.L.

\section*{FET APPLICATIONS HANDBOOK}

\section*{Edited by Jerome Eimbinder}

Published by Foulsham-Tab
350 pages, \(8 \frac{1}{2}\) in \(\times 5 \frac{1}{2}\) in. Price \(£ 1.85\)

THIS book is a collection of applications data from a wide number of manufacturers and provides an excellent reference work for anyone who is involved in the use of field effect transistors (f.e.t.s).
All types of f.e.t.s are described from the simple junction types to the complementary M.O.S. systems used in the pocket calculators of recent years. Many applications of f.e.t.s are described ranging from voltage controlled resistors to amplifiers, switches and digital circuits.
S.R.L.


By Brian Baily．．．

\author{
＂．．．more things in heaven
} and earth．．．＂

The first reaction of many readers when they read the above title was probably to ask＂What＇s a subject like this doing in P．E．？＂You may well ask．Others may have taken the view that our friend Ed is off his rocker and yet others may take the opposite view，and exclaim ＂At last Ed has come to his（extra） senses，and allowed publication of this previously＇tabooed＇subject．＂｜

The fact is，this article was written long before Uri Geller＇s impressive fork－（if not mind）－ bending demonstration on BBC tele－ vision，but friend Ed was a little sceptical I think before that night when I talked him into viewing it！ We must all be grateful for his open－minded approach in allowing this article to appear．
＇Naturally，＇there will be those for and those against，but whether you are a sheep or a goat（these are the two rather inappropriate terms that sceptics use to describe the believers and non－believers，re－ spectively）there is always a good reason to give the whole matter an－ other look．Remember，times change，and so do peoples＇ideas， evidence，equipment，and even phenomena．

First，let us get quite clear what is meant by extra－sensory percep－ tion（E．S．P．）．Take the words liter－ ally，and you will get the best definition possible．They mean per－ ception of events（etc．）by some means not obviously connected with any of the five normal senses．

Let me hasten at this point to make a quick distinction or two． Unfortunately，the Press，Radio，and TV media often treat all subjects
which cannot be explained by pre－ sently－accepted terms of analysis－ witcheraft，black magic，ghosts ．．． what－have－you－as E．S．P．This is a great pity，because there are quite different forms of unexplained phenomena，and the only true umbrella titles which could be fairly used to embrace the lot，would be＂the occult＂（unknown），or ＂paraphysics．＂

However，when writing a subject which must be sensational，so－called evidence is often dragged in from as many different fields as possible to try and get the reader all excited and on the edge of his chair．In fact，the opposite often happens，and many are left believing less one way or the other than when they first started reading！
So，though other subjects come under the above general headings， they are not all E．S．P．Such sub－ jects will be described in this series as the etceteras．

\section*{Reasonable Approach}

It must always be remembered that there is no commonplace thing of today that，at some time past， was not considered－impossible．We are all still learning，and humility is，to the truly rationally－minded， something which can only in－ crease，not decrease，with in－ creased knowledge．Just because humans have relied on speech，sight， etc．，for communication and direct physical intervention for getting things done，is not to say that there is no need of a set of equally valid alternatives in his make－up，which needs only to be developed，once discovered－always given he adopts a reasonable approach．
あ あ 亚

E．S．P．is nothing new．One can go back to the New Testament，if this is considered far enough，and find references to＂gifts of the spirit＂．I will not digress further， other than to say that evidence of the possibility of E．S．P．has cropped up probably as long as man has inhabited the earth，and in other forms，before．

E．S．P．rarely occurs under con－ trolled laboratory conditions．Read a person＇s thoughts say，once，and then try to continue the pattern under strict control，and the chances are that you will not succeed again even once．This is the very thing the sceptic looks for，and when it occurs he jumps in with both feet and says＂Pooh what did I tell

you？＂Ok，ok，I say，I am as aware as you that the effect is not re－ peated under these conditions，bui might this not be a point worthy of investigation？＂Why？＂he would say，and then would come my first analogy，speaking as an electronics bod．

Take out your comb and comt your hair，or if this is greasy，rub it on your Harris Tweed jacket．The comb is now charged，and you can prove this by picking up small bits of tissue paper from the table． What voltage？Would you believe there＇s probably more than 5 kV on it？There probably is！

Now ask the sceptic to get his voltmeter and place one lead on the gas or water pipe（earth）and the other on the comb and see what he gets．Tell him the meter is at least 20 kilohms per volt．Naturally， he gets no reading．Why，you were on the 250 V scale and the needle didn＇t even flick．Reason is that the meter requires current to flow to operate its needle，and the amount of current required for this is in excess of what the comb could supply despite over 5 kV on it，at the point of contact with the meter lead．

Could E．S．P．be elusive for some similar reason？By this I mean， by trying to＂tap＂its power to prove an experiment，could we re－ quire too much＂current＂for the source to supply and operate our ＂meter＂？Perhaps the tension caused by the anticipation of all spectators present draws off the very power we are trying to use！

Next month：The Scientists in－ vestigate．

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ELEOTROALUE Catalogue 7
}

\section*{An \(\mathbf{A}\) to \(\mathbf{Z}\) guide to component buying}

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3015 G showing + or - and 1 and dec. pt, 41-20.

\section*{DALY ELECTROLYTIC CAPACITORS} in cons, plostic sleeved
\(1,000 \mathrm{mF} / 25 \mathrm{~V} .28 \mathrm{p} ; 2,000 \mathrm{mF} / 25 \mathrm{~V}, 37 \mathrm{p} ; 5.000 \mathrm{mF}\) \(25 \mathrm{~V}, 62 \mathrm{p} ; 1,000 \mathrm{mF} / 50 \mathrm{~V}, 41 \mathrm{p} ; 2,000 \mathrm{mF} / 50 \mathrm{~V}, 57 \mathrm{p}\) \(5,000 \mathrm{mF} / 50 \mathrm{~V}, ~ £ 1 \cdot 18 ; 1,000 \mathrm{mF} / 100 \mathrm{~V}, 82 \mathrm{p} ; 2,200 \mathrm{mF}\)


POLYESTER CAPACITORS TYPE C280
Radial leads for P.C.B. mounting. Working voltoge \(0.01,0.0+5,0.022,0.033,0.047,3 p\) each: 0.068 \(0.1,0.15 .4 p\) each; \(0.22,5 p ; 0.33,7 p ; 0.47,8 p\) \(0.68,11 \mathrm{p} ; 1 \cdot 0,14 \mathrm{p}\); \(1 \cdot 5,21 \mathrm{p} ; 2 \cdot 2,24 \mathrm{p}\).
SILVERED MICA CAPACITORS
Working voltage 500 V d.e.
Values in \(\mathrm{PF}-2.2\) to 820 in 32 stages, \(6 p\) each 1,000. ,600, 12p each: 1,800, 8p: 2,200 10p 6,80020 p ; 8, 200, 10,000 25p each.

\section*{POTENTIOMETERS}

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Double wipers for good contact and long warking life P20 SINGLE linear 100 ohms to \(2 \cdot 2 \mathrm{M} \Omega\), I4p each P20 SINGLE log. \(47 \mathrm{k} \Omega\) to 22 Mn , 14 p each.
\(\mathrm{JP20}\) DUAL GANG linear \(4.7 \mathrm{k} \Omega\) to \(2.2 \mathrm{M} \Omega 4 \mathrm{p}\) jP20 DUAL GANG log. \(4 \cdot 7 \mathrm{k} \Omega\) to \(2 \cdot 2 \mathrm{Mn} 48 \mathrm{p}\) JP20 DUAL GANG log/antilog 10L, 22K, 47 K IMn only, 48p each.
JP2O DUAL GANG antilog lok only 48 p
\(2 A D D^{P}\) mains switch with any of above \(14 p\) extra. Decades of 10,22 and 47 only avaitable in ranges above. SKELETON CARBON PRESETS 6p each
SLIDER POTENTIOMETERS
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each.
Control knobs blk/wht/red/yel/grn/blue/dk. grey/i grey, 7p each

\section*{ELECTROLYTIC CAPACITORS}

Axial Lead
\begin{tabular}{|c|c|c|c|c|c|c|c|c|}
\hline uF & \[
3 V
\] & 6.3 V & 10 V & 16 V & 25 V & 40 V & 63 V & 100 V \\
\hline 0.47 & - & & - & & & & Ilp & 8 p \\
\hline \(1 \cdot 0\) & & & & & & \(11 p\) & & 8p \\
\hline 2.2 & - & & & & 11p & & \({ }^{8 p}\) & 9 9 \\
\hline 4.7 & - & 二 & - & 11p & & 8 p & 9 p & 8 p \\
\hline 10 & - & - & - & - & 8 p & 9 p & 8 p & 8p \\
\hline 22 & - & - & 8 p & - & 9 p & 8 p & 8 p & 10p \\
\hline 47 & 8 p & - & 9 p & 8 8p & 9 p & 8 p & 10p & \(13 p\) \\
\hline 100 & \({ }^{9 p}\) & 8p & 8 p & 8 p & 9p & 10p & 12p & 19p \\
\hline 220 & 8 p & 8 p & 9p & 10p & 10p & \(11 p\) & 17p & 28p \\
\hline 470 & \(9 p\) & 10p & 10p & \(11 p\) & 13p & 17p & 24p & 45p \\
\hline 1.000 & 11p & 13p & 13 p & 17p & 20p & 25p & \(41 p\) & - \\
\hline 2,200 & 15p & 18p & 23p & 26p & 37p & \(41 p\) & - & - \\
\hline 4,700 & 26p & 30p & 19p & 44p & 58p & & & \\
\hline
\end{tabular}

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Code Watts ohms \(1-9 \quad 10-99 \quad 100\) up C \(\quad 1 / 20 \quad 82-220 \mathrm{~K} \quad 9 \quad \begin{gathered}\text { (see note below) } \\ \text { C }\end{gathered}\)
\begin{tabular}{llcllll}
\(C\) & \(1 / 20\) & \(82-220 K\) & (see note below) & \\
\(C\) & \(4.7-470 K\) & 1.3 & 1.1 & 0.9 & nett \\
\(C\) & 3 & \(4.7-10 M\) & 1.3 & 1.1 & 0.9 & nett \\
\(C\) & 1 & \(4.7-10 M\) & 1.5 & 1.2 & 0.97 & nett \\
\(C\) & 1 & \(4.7-10 M\) & 3.2 & 2.5 & 1.92 & nett \\
\(C\) & 1 & \(10-1 M\) & 4 & 3.3 & 2.3 & nett \\
MO & & \(0.22-3.9\) & 7 & 7 & 8 & \\
\(W W\) & 1 & \(1-10 K\) & 7 & 7 & 8 & \\
\(W W\) & 3 & \(1-10 K\) & 9 & 9 & 8 &
\end{tabular}

Codes: C-carbon film, high stability, low noise. MO-metal oxide. Electrosil TR5, ultra low noise WW-wire wound, Plessey
Valuen: All El2 except CiW, CiW and MOtW at Ei \(2,10,12,15,18,22,27,33,39,47,56,68,82\). E12, \(10,12,15,18,22,27,33,39,4,56,68,821,62\)
E24 as El2 plus \(14,13,16,20,24,30,36,43,51,62\) 75,91 and their decades. 20 at \(2 \%\) and WW \(10 \% \pm 0.05 \Omega\)
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\(2-7 V\) to 36 V , I4p each; \(1 \mathrm{~W}: 6.8 \mathrm{~V}\) to 82 V , 21 p each \(2-7 \mathrm{~V}\) to 36 V , \(14 p\) each: \(1 \mathrm{~W}: 6.8 \mathrm{~V}\) to \(82 \mathrm{~V}, 21 \mathrm{p}\) each;
\(1.5 \mathrm{~W}: 4.7 \mathrm{~V}\) to \(75 \mathrm{~V}, 48 \mathrm{p}\) each. Clip to increase 1.5 W rating to 3 watts (type 266F) 4p.

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Savbit alloy, \(81 \cdot 16=\square\)

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\section*{beating the clock}

Coming into service in a few weeks is Cantat 2 , the 2,805 -mile long transatlantic cable costing \(£ 30\) million and capable of handling 1,840 telephone conversations simultaneously.

Laying of the cable had its share of drama. The mini-submarine Pisces 3 sank while burying part of the cable on the UK continental shelf. The two-man crew were rescued, happily, after three days of being trapped on the sea-bed. The craft was also recovered and will be in service again soon.

The cable-laying ship Mercury was also damaged in a storm of force 11, forcing her to operate for a time on emergency power. But despite all the upsets, the cable-laying was completed six days ahead of schedule which makes quite a change in a world of go-slow or stop.

The other heartening news out of the Cantat 2 saga is that because of the vastly increased number of channels available, the new cable is 20 times cheaper per circuit than the transatlantic cable laid between the UK and USA in 1956. But don't expect the cost of a call to go down.

\section*{HI-FI HIT}

Hi-Fi, like the motor car, has become the index of a nation's prosperity. But what was astonishing in the good old days of 1973 was the number of equipments advertised to the general public-not the addict-at prices of \(£ 200\) up, and reportedly selling well. Top weight equipment for the real enthusiast was typically Trio's Package 10
quadraphonic package listed by Lasky's at \(£ 460\) cash or \(£ 577 \cdot 48\) on HP. At the bottom of the pack were plenty of modest units around £30.
A survey of the audio market "Finresearch Audio Industry Service" published by Ovum Ltd. at £30 lists 380 companies active in the audio business which has trebled in size in three years. Such was the boom, according to editor Tony Beaumont, that three new companies have been entering the field every week for the past two years.

First 'audio separates' | recall was back in 1937 when I was enchanted with the reproduction of 78 r.p.m. discs on the famous corner horn loudspeaker developed by P. A. H. G. Voigt. But I see Ferguson claiming they "invented" what they call Unit Audio just 30 years later in 1967. Claims and counter-claims in the colour ads are numerous and often funny.

This clearly will not be a good year for the audio business. HP restrictions and a general tightening of belts will see to that. But the audio scene will not collapse. It'll still be comparatively big-biz as well as show-biz.

\section*{MAINTENANCE PROBLEM}

Try to imagine 15,000 fixed transmitter/receivers of powers from five to 300 W and 25,000 mobile vehicle-mounted transmitter/receivers. Tack on 30,000 personal portable transmitter/receivers and 200,000 nickel/cadmium batteries to power them. And just add to that little lot 15,000 pocket alarm receivers and all their batteries.

In round figures, this is the size of the maintenance problem of the Home Office Directorate of Telecommunications which looks after the procurement and servicing of Britain's police and firefighting networks as well as a few others. The problem was discussed in some detail by A. N. Holdstock in a paper given at "Automatic Testing 73', the first full-scale exhibition and conference to be held on the subject. The magnitude of the Home Office's task was such that automatic testing was seen as a solution as early as 1967. Holdstock reports several equipments now in full service but reports that no radio equipment manufacturers are yet designing sets with in-field maintenance by ATE as a prime consideration. Consequently, there are always snags in having to build special jigs and adaptors before ATE can be successfully used in the field.

The next big job is provision of ATE for 1,000 alpha-numeric videodisplay terminals now being fitted
in police stations throughout the country. Each of them has six large PCBs and 530 TTL integrated circuits.
A new and profitable outlet for ATE on production lines is in the automotive industry. No less than six papers were read on the subject including one on the experience of Ford in checking out car electrical systems. Ford will, in fact, move 100 per cent to ATE for this purpose in the UK during this year.

So successful was "Automatic Testing 73." that organiser John Shearsmith has already announced November 5, 1974, as opening day for the next show. The first exhibition attracted 50 companies and 4,000 visitors. Seven hundred delegates attended the technical conference.

\section*{INNOVATION}

I have to confess that I was one of the sceptics when Advance Electronics went into the calculator business in mid-1972. Like many others I had underrated their ingenuity.

Advance didn't take the easy path by buying off-the-shelf LSI devices. They designed their own in conjunction with General Instrument Microelectronics and did it in such a way that they could be used in many other applications. So watch out for some competitive lines from Advance in business machines.

\section*{MINICOMPUTER LEAGUE}

A league table of minicomputer manufacturers prepared by Modern Data in the USA shows Digital Equipment Corporation still way ahead with 16,473 units shipped in 1973. Runner up is Data General with 4,230 and Hewlett Packard in third spot has a score of 2,855, marginally ahead of Varian with 2,337. Honeywell ranks only No. 8 with 1,564 while Texas Instruments, mighty in other fields, is bottom and unlucky thirteenth with 452.

Computer Automation, who came to the market with the concept of the Naked Mini, ranks fifth with 1,915 units shipped in the year. Stripped of all frills, the Naked Mini was offered at a rock-bottom price. Since then CA has offered the "Computer-on-a-card" with the whole package on a single printed circuit board, an even lower priced unit for people who want to build a computer into their own equipment.

Digital Equipment still has the stranglehold in minis that IBM has in the larger machine market. But watch out for increased competition and a re-shuffling of positions lower down the table.

\title{
PRTENTE RETOECNO
}

\section*{AUTOMATIC COMMENTARIES}

The Secretary of State for Defence has a new British patent (BP 1325 161) which concerns automatic machines for playing commentaries to accompany exhibits in museums or exhibitions.

The object of the invention is to provide a replay device which works on the closed loop principle and will always be at the correct start point when brought into action by a new listener. This is relatively easy to achieve if the loop is only to play once, but much more difficult to achieve if the listener is to be afforded the opportunity of
when the earpiece is lifted off its rest. The switch is open while the earpiece is on its rest. Amplifier 1 is coupled to the headset earpiece through transformer T1.

When a would-be listener picks up the earpiece \(S 1\) closes and the Schmitt trigger changes state to start the tape transport motor. The audio frequencies from the tape are reproduced through the earpiece and used, after rectification, to hold the Schmitt in its active state. This holding action will continue even when the earpiece is returned to its rest.

However, when the audio signals cease, the voltage across C1 dies away and the Schmitt trigger reverts to its former state and the

listening to the loop more than once.

The tape replay device works with an endless loop of magnetic tape and its audio output passes through the pre-amplifier to two amplifiers, see Fig. 1. The output from amplifier 2 is fed to a simple rectifier circuit, D1 and C1, and passed via R1 to a Schmitt trigger circuit, TR1, TR2. The Schmitt trigger is connected to the power supply and controls operation of the tape replay motor.

The switch S1 is mounted as part of a telephone handset and closes
tape motor stops. Thus the loop stops at its start position. The decay time constant of the voltage across C1 must, of course, be long enough to tide over natural pauses in the commentary.

If the listener wishes to hear the commentary through again, then he keeps the earpiece off its rest; the switch S1 holds the trigger circuit active and the loop is replayed.

The patent specification also details a twin channel system which can be used to control visual displays.

\section*{SCORING FOR BOXING AND OTHER SPORTS BP 1325644}

Although BP 1325644 from E. J. Sweeny of Ohio, USA, is intended primarily for the scoring of boxing matches by judges, its technology will find wider uses.

The specific intention is to enable three judges to score independently with the sum of their points displayed on a common board.
A master console and three identical scoring booths are spaced round a boxing ring with a visual score board mounted above. A single point will be recorded on the score board only after the master console has received a multiple number of point pulses from the individual scoring booths, e.g. three point pulses must be accumulated from one or more booths before a single point is recorded on the board.
The controls within the booths are duplicated for the two fighters.

\section*{VEHICLE GUIDANCE BP 1320314}

Interest appears to be growing in the technology of automatic vehicle guidance. In BP 1320314 Autotrack Systems Ltd., and P.C.D. Ltd., jointly describe the automatic remote control of a tractor ploughing a field. The principle of the invention, however, is much more widely applicable, e.g. garden lawn mowing to model car control.

Guide wires are buried half a metre deep and six metres apart along the routes which the vehicle should take. These guides carry a continuous tone current of 2 kHz and the a.c. field that this produces is detected by a detector head on the vehicle.

The detector head has a vertical pick-up coil and a horizontal pickup coil. With the head exactly above the guide wire there is no phase difference between the outputs of the two coils, but if the head is moved to one side a phase difference appears in the appropriate sense and of magnitude proportional to the distance moved.

The two coils in the patented circuit control solenoids for left and right turn and operate pneumatic, hydraulic or electrical steering gear on board the tractor.

\title{
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\section*{semiconductors}

\begin{tabular}{|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{TRIACS} & \multicolumn{2}{|l|}{\multirow[t]{2}{*}{LINEAR OP. AMPS}} & \multicolumn{2}{|l|}{\begin{tabular}{l}
ZENER OIODES \\
BZY88 series 400 mW
\end{tabular}} \\
\hline SC40D 6A
SC40E 64
500 V & \(\Sigma 1.00\)
\(\Sigma 1.20\) & & & 15 Watt range
10 Watt range & 45p \\
\hline SC45D 10A 400 V & £1.25 & 709C T099]IL & 40 p & TIL 209 H P L.E.D. & \\
\hline SC45E 10A 500V & 21.45 & 723 C TO99/DAL & E1.05 & THL 209 H P 5082 & 280 \\
\hline SC50D 15A 500V & 81.85 & 741 C OL
747 C TO99/OL & 81.10 & ORP 12 L.O.R. & 50p \\
\hline \multirow[t]{3}{*}{DIAC} & \multirow[t]{2}{*}{25p} & 72748 P DIL

723 C TO99 & 85p
c1.05 & & \\
\hline & & 723C TO93
CA3014 & c1. 55 &  & \\
\hline \multicolumn{2}{|l|}{\multirow[b]{2}{*}{S.C.R.}} & CA3018 & 51.00 & & \\
\hline & & CA3026
CA3036 & \$1.20 & \multicolumn{2}{|l|}{Cossor-1035 DE. 1049 Mk .3 A DB. 1049 Mk .4 . D B.} \\
\hline \multicolumn{2}{|l|}{BTY 79/400R} & CA 3046 & 95p & \multicolumn{2}{|l|}{EM1-WM2 WM3B WM7 WM8} \\
\hline \multicolumn{2}{|l|}{CRS 1/05 25p} & \multirow[t]{2}{*}{CA3090} & ca. & WM16 WM 18 WM 26. & \\
\hline CRS \(1 / 10\) & \(30 p\) & \multicolumn{2}{|l|}{\multirow[t]{2}{*}{MC1303L 51.85}} & \multicolumn{2}{|l|}{Solartron-AD557 CD5132 CD523} \\
\hline CRS 1/20 & 30p & & & \multicolumn{2}{|l|}{S.2. CDices and details on application.} \\
\hline CRS 1/40 & 35p & MFC4000B 60p & 60 & \multicolumn{2}{|l|}{\multirow[t]{2}{*}{phone 01-994 6275. Inspection wel-}} \\
\hline CRS 31035 & 30 p & \multicolumn{2}{|l|}{SL403D} & & \\
\hline CRS \(3 / 10\) & 30 p & \multicolumn{2}{|l|}{\(\begin{array}{ll}\text { ZNA14 } \\ \text { LM309K } & 51.25 \\ \text { ¢1. } 20\end{array}\)} & \multicolumn{2}{|l|}{} \\
\hline CRS 3/40 & \(50 p\) & TAD100 & ع1. 00 & \multicolumn{2}{|l|}{BRIDGE RECTIFIERS} \\
\hline \multicolumn{2}{|l|}{CRS 3.60 80p} & \multicolumn{2}{|l|}{TOKO FILTER 45p} & \multicolumn{2}{|l|}{} \\
\hline CRS 7/400 & 700 & TAD110 & [2.00 & BY164 1.4A 200 V & 57p \\
\hline CRS 7/600 & 90p & & & CIC2-100 2A, 100 V & \\
\hline CRS 16/100 & 85p & & & 6 A .100 V & \\
\hline CRS 16/200 & 720 & & & & \\
\hline \multicolumn{2}{|l|}{CRS 16/600 \(\quad\) E1.05} & \multicolumn{4}{|l|}{\multirow[t]{2}{*}{V.A.T.}} \\
\hline C 106 D & 65p & & & & \\
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& \text { 16RC/20 } \\
& \text { 2N3525 }
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\begin{array}{r}
700 \\
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\begin{tabular}{|c|c|c|}
\hline \begin{tabular}{l}
\[
\begin{aligned}
& \text { SGS } \\
& \text { EA1000 } \\
& \text { 3W } \\
& \text { AMPLIFIER } \\
& \text { MODULE } \\
& \mathbf{E 2 . 4 5}
\end{aligned}
\] \\
HANDBOOK 10p \\
Solldev 15W, Clase 8 Audlo Amp, Thlek Film Circuit, \(\{3.65\); Handbook. 15p. tBAB00 5W Audio I.C. 51-50; Data. 10p. \\
UA041 5W Amplifier Module. E3.75
\end{tabular} & \begin{tabular}{l}
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4250 Polyester. Polycarbonate. densers. \\
525 Potentiometers. assorted \\
6250 High-stab. 1\%. \(2 \% .5 \%\) resis \\
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8 11b Assorted nuts, Dolts, washer \\
925 Assorted switches, rotary, leve etc. \\
1050 Preset Potentiometers. \\
ALL COMPONENTS NEW AND UN p.p. per pack. \(\mathbf{5} 5\) for 5 pac
\end{tabular} & \begin{tabular}{l}
Con- \\
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\(+25 p\)
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\hline \multicolumn{3}{|c|}{TEST EQUIPMENT} \\
\hline TEKTRONIX Type 1 H P Transter Oscilla H P Sweep Oscillato H P Sweep Os cillat MARCONIVHF-SHF & Pulse Generator \(540 \mathrm{~B} \cdot 100-200 \mathrm{MHz}\) 101-686C. 7-11 GHz \(684 \mathrm{C} .4-8 \mathrm{GHz}\) nal Generator. TF 1058 & 175
\(\$ 100\)
\(\$ 120\)
\(£ 150\)
\(E 200\) \\
\hline PYE LING Automatic MARCONI TF 1400 CINTEL 1873 Square & XLYSTON POWER SUPPLIES rge selection in stock. e20 each. yeling Oscillator. A.C. \(0 \cdot 1.5 \mathrm{~Hz}-5 \mathrm{~L} \mathrm{~Hz}\) ble Puise Generator ave and Pulse Generator & 550
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550 \\
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\section*{SWITCHES}

DP／IOT Toggle 25p \＄p／8T Toggle 18p

\section*{FUSES}
 4IICK－bLow 4pea．ANTI－sURGE 5pea

\section*{EARPHONES}

8 ohull \(\%\) mm plog 22
8 ohuls 3 ．Јura plug 220
DYNAMIC MICROPHONES

3－WAY STEREO HEAD－
PHONE JUNCTION BOX
\(\frac{\text { H1012 }}{\text { 2－WAY CROSSOVER }}\)
NETWORK
K 4007 ． 80 ohsur luap．Insertion toss 3 dB £1－21
SELECTED RESISTORS
R1 Our mixi，i watt carbon－ 50 for 550
CAR STEREO SPEAKERS
（Angied） \(\mathbf{£ 3 . 8 5}\) per pal
BI－PAK
CATALOGUE AND LISTS Send S．A．E．and 10p．
INSTRUMENT CASES

\begin{tabular}{|c|c|c|c|c|c|c|}
\hline \multicolumn{2}{|l|}{\multirow[t]{2}{*}{No．Length}} & \multicolumn{2}{|r|}{Width} & \multicolumn{2}{|l|}{Height} & \multirow[t]{3}{*}{\[
\begin{aligned}
& \text { Price } \\
& 90 p \\
& \mathbf{9 1 . 2 0}
\end{aligned}
\]} \\
\hline & & \(\times\) & J＂ & \(\times\) & & \\
\hline 13 V 2 & \(11^{*}\) & \(\times\) & \(6^{*}\) & \(\times\) & \(3^{*}\) & \\
\hline \multicolumn{7}{|l|}{ALUMINIUM BOXES} \\
\hline BA！ & 5！ & \(\times\) & 2 & & \(11^{-}\) & 42p \\
\hline BAZ & 4 － & \(\times\) & 4 & \(\times\) & \(11^{\circ}\) & 41 D \\
\hline \(\mathrm{BA}_{3}\) & \({ }^{*}\) & \(\times\) & \(3{ }^{-1}\) & \(\times\) & 1. & 41p \\
\hline BA4 & \(5{ }^{*}\) & \(\times\) & \(4{ }^{*}\) & \(\times\) & \(1{ }^{*}\) & 47p \\
\hline BAE & \({ }^{\text {a }}\) & \(\stackrel{x}{x}\) & \({ }^{2 / 8}\) & \(\times\) & \({ }^{2}\) & 41 p \\
\hline BAT & \(\stackrel{*}{*}\) & \(\times\) & \(\stackrel{5}{5}\) & & \({ }^{1}{ }^{-}\) & \({ }_{868} 8\) \\
\hline Bas & \(8^{8}\) & \(\times\) & \({ }^{*}\) & \(\times\) & 3. & 840 \\
\hline BA9 & \(6^{*}\) & \(\times\) & \(4 *\) & \(\times\) & 2＊ & 84 p \\
\hline
\end{tabular}

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Chrome Finish Model \(60 £ 1\)－50

Ref．36A．Recoril／Stylus Cleaning Kit 28p
Ref．43．Record Care Kit \(£ 2.35\)
Ref．31．Cassette Heall Cleaner 54p
Ref．32．Tape editing Kit \(£ 1.54\)
Sollel 9．Wire Stripper／Cutter 83D
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X \(2 \overline{5} .2 \overline{5}\) watt \(£ 1.93\)
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\section*{ANTEX BITS and ELEMENTS}

\section*{Bits N}

102 For model CN240 \(\frac{3}{3}\) ． 102 For model CN240 id 100 For model CCN 240 素 101 For model CCN240 \({ }^{*}\)＂ 102 For model CCN240 ！＊ 1020 For model（i2．40 素＂ 021 For model G240 \(n_{n}^{\prime \prime}\) 1022 For model G240 18

\(\therefore 1\) For model X25 ：
ELEMENTS
ECN 240 21－16
ECCN 240 £1－32
E（： 240 \＄1．16
EX 25 \(21-16\)

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No．Qty

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count by weight count by weight
Precision Resistor mixell valnes
dth W F values \(\qquad\) Tuning Gange MW／LW VHI 0.55 colours
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C． 3 Micro Suitches
C10 15 Assorted Pots \＆Pre－Seta


Jack Socket． \(3 \times 3\) ． 8tandard Switch Type C12 \(40 \begin{gathered}\text { Paper Condense } \\ \text { mixed values }\end{gathered}\)
C13 20 Flectrolytics Trans．typer C14 1 Pack assorted Hardware－ Mains Slide 8 witches C16 20 Assorted Tag 8trips \＆Panels C17 10 Assorted Control Knobs C18 4 Rotary Wave Change Switches 0．55 C19 3 Relays 6－24V Operating C2O 4

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\(10^{\prime \prime} \times 5 \%\)
0.55

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Ret． 32 A ．Stylus Balance \(£ 1.36\) Ref．J．Tape Head Cleaning Kit 51p Ref．34．Casset te Case \(£ 1.27\) Ref．J6．Hi－FiStereo Hints \＆Tijs 32p

\section*{PLUGS AND SOCKETS} SOCEETS
PS 35 DIN 2 Pin（Speaker） P＇ 37 DIN os Pin \(180^{\circ}\) PS 38 DIN 5 Pin \(240^{\circ}\) PS 39 Jack 2＇5mm Swltehed PS 40 Jack 3 onnmin Switched \(\begin{array}{ll}\text { PS } 41 & \text { Jack } f^{*} \text { Suitched } \\ \text { PS } 40 & \text { Jack Stereo Swit }\end{array}\) PS 49
PS 43
Jack Stereo Switched
Phono Single phono Single ＇hono Double Car Aerial \(\begin{array}{ll}\text { PG } 46 & \text { Co－Axial Burface } \\ \text { PG } 47 & \text { Co－Axial Fliah }\end{array}\)

INLINE SOCKETS
PS 21 D．I．N． 2 Pin（Speake！） P＇S 22 D．I．N． 3 Pin
PS：33 D．1．N．J Pin \(180^{\circ}\)
\(\begin{array}{ll}\text { PS } 24 & \text { D．I．N．JPin } 240 \\ \text { PS } 25 & \text { Jack } 2.5 \mathrm{mmm} \text { Platic }\end{array}\) \(\begin{array}{ll}\text { PS } 25 & \text { Jack } 2 \text {＇amm Plastic } \\ \text { PS } 20 & \text { Jack } 3 \cdot \text { bnum Plastic }\end{array}\) PS 27 Jach 1＂Plastic PS 28 Jack ！＊Screened PS 29 Jack Stereo Plastic
PS 30 Jack Stereo Screened PS 31 Phono Screened PS 3：Car Aerial PS 33 Co－Axia！

\section*{PLUGS}

PB 1 D．1．N． 2 Pin（8peaker）

\section*{D．IN． 3 l＇in}

PS 3 I．I．N． 4 Pin
PS 4 D．I．N．\(\overline{0}\) Pin \(180^{\circ}\)
D．I．N．\(\overline{5} \operatorname{Pin} 240^{\circ}\)
D．IN． 6 Pin
Pg 8 Jack 2 ómm Screened P＇s 9 Jack 3 ofmm Plastic
PS 10 Jack 3．Jum Bcreened
P＇ 11 Jack J＂Plastic
PS 1：2 Jack ：Bcreened
PS 18 Jack Btereo Bcreened
PS 14 Phono
PS 1．̄ Car Aeria
PS 16 Co－Axial

\section*{CABLES}

Single Lapped Screen
CP Twin Common Screen CP 3 Stereo Screened
CP 4 Fons Core Conmon Bereen
Four Core Indivilually Screened 0.30 Microphone Fuily Braided Cable 0.10 Three Core Mains Cable 0.07 CP 7 Three Core Mams Cable CP 8 Twin Oval Main CP 10 Low Loss Co－Axial

\section*{CARBON}

\section*{POTENTIOMETERS}

Log and Lin
\(4.7 \mathrm{~K}, 10 \mathrm{~K}, 22 \mathrm{~K}, 47 \mathrm{~K}, 100 \mathrm{~K}, 220 \mathrm{~K}, 470 \mathrm{~K}\) 1M，em
vC1 Single Less 8 witch
VC 2 Single D．P．Suitch CC 3 Tandern Less switch 1 C 41 K Lin Less switch － 100 K Log anti－Log

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anodifications
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and its construction
\(\mathbf{2 0 p}\)
Reactance－Frequency Chart for Aulio \＆RF use
Resistor Colour Code Disc Calculator

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ACOB GP91－18C． 200 mV at \(1.2 \mathrm{~cm} / \mathrm{sec} 21.16\) ACOS（iP93－1．280mV at \(1 \mathrm{cmfacc} \quad\) \＆1．65
 TTC J－2005．Crystal／Hi Output
\(\qquad\)
TTC J 200 Cs stereo／Hi Output \(\quad \Sigma 1.80\)
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CARBON FILM RESISTORS
The Ely Range of Carbon Film Resistors， I Hat atallable PAKS of piece R1 10 Mixed 100 ohms－ 820 ohms R2 50 Mixed 1 K ohms－8．2K ohms 40p R3 50 Mixed 10 K ohms－82K ohnıs 40p R4 50 Mixed 100 K ohms． 1 Meg．ohms 40p THESE ARE UNBEATABLE PRICES－ LESS THAN 1P EACH INCL．V．A．T

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The AL10, AL20 abit AL30 units ar similar in their appearsnce and In their general specification. However, careful selection of the plastic power devices has esulted in a range of output lowers from to 10 watts R.M.S
The versatility of their design makes them itheal for une in record players, tape recorders, stereo amplifers and casmette and cartridge tape players in the car and at home
\begin{tabular}{|c|c|c|}
\hline Parameter & Condition: & Pertormance \\
\hline HARMONIC DISTORTION & Po \(=3 \mathrm{WATTS} \mathrm{f}=1 \mathrm{KHz}\) & 0.25\% \\
\hline LOAD IMPEDANCE & - & 8-16n \\
\hline INIUT IMPEDANCE & \(\mathrm{f}=1 \mathrm{kHz}\) & 100 kn \\
\hline FREQUENCY RESPONSE - 3413 & \(\mathbf{P o}=2 \mathrm{~W}\) atts & i0 Hz-25K Hz \\
\hline gensitivity for rated o/P & \(\mathrm{V}_{5}=2 \overline{\mathrm{~J}} \mathrm{~V} \cdot \mathrm{Rl}=8 \Omega \mathrm{t}=1 \mathrm{KHz}\) & 75 mV , RMs \\
\hline DIMENSIONS & - & \(3^{*} \times 2 \frac{1}{3}^{*}=1^{\circ}\) \\
\hline
\end{tabular}

The above table relates to the AL10. AL20 and AL30 in their working conditions.
\begin{tabular}{|c|}
\hline Parameter \\
\hline Maximum Supply Voltage \\
\hline Power out for \(2 \%\) T.I. B
\[
(\mathrm{KL}=8 \Omega \mathrm{f}=1 \mathrm{KHz})
\] \\
\hline
\end{tabular}

AUDIO AMPLIFIER MODULES
\(\begin{array}{ll}\text { AL 10. } & 3 \text { watts } \\ \text { AL } 20 . & 5 \\ \text { wratts }\end{array}\)

POWER SUPPLIES
PG12. (Use with AL10 \& AL:20) 88p SPM 80. (1 We with also AL30 \& Als0) FRONT PANELS PA 12 with Knohs

\section*{\(0.1 \%\) DISTORTION!}

HI-FI AUDIO AMPLIFIER

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\(\star\) Frequency Response 15 Hz to \(100,000-1 \mathrm{~dB}\).
* Load-3, 4, 8 or 16 ohms.
* Distortion-better than \(1 \%\) at 1 KHz .
* Signal to noise ratio 80 dB .

ONLY

\(\star\) Supply voltage \(10-35\) Volts.
* Overall size 63 mm \(105 \mathrm{~mm} \times 13 \mathrm{~mm}\).

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Three switched stereo inputa, and rumble and scratch fliters are features of the PA100, which also has a STEREO/MONO switch, volume, balance and contlnuously varimble hass and treble controls.


SPECIFICATION
Frequency Response Harmonic Itistortion Inputa: 1. Tape Head Radio, Tuner
All input voltages are for an output of 250 mV . Tape and P.U. imputa All input voltages are for an output of 250 mV . Tape and P. U. input 1 ass Control \(\pm 15 \mathrm{~dB}\) at 20 Hz Treble Control
Filters: Rumble ( H igh Pass) Scratch (Low Pass)
Signal/Noise Rallo
Input overload
Dupplysions

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better then - BodB
\(+26 \mathrm{~dB}\)
\(292 \mathrm{~mm} \times 82 \mathrm{~m}\)
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ranges. 20.000opy. rangess. 20.0000 pr .
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 AC. 3/10/50/250/ current 10/100uA/10/ 10/100/500mA/2.5/100A. Resistence:
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3.0
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KAMODEN 72.200 Multitester
High sansitivity
tenier. 200,000 opy
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Sine \(18 \times 200,000\)
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Buit-in timb base Calibrator and amplitud Callbrator. Supplied complere with all sccossorias OUR PRICE \(\mathbf{£ 8 7 . 0 0}\) Carr. pard MCA220 Automatic Voltage Stabiliser
Stabiliser
Input \(88-125 \mathrm{~V}\)
AC or \(176-250\)
AC Outpu
120 AC or
240 ACC
200 A Arsing.

OUR PRICE E11.97 P\&P 50p
PS100B Regulated POWER SUPPLY UNIT


CLEAR PLASTI
Size: \(100 \times 80 \mathrm{~mm}\)
\begin{tabular}{|c|c|c|c|c|}
\hline 50uA & E4.55 & \multicolumn{3}{|l|}{} \\
\hline 100UA .. .. & ¢4.35 & \multicolumn{3}{|c|}{\multirow{4}{*}{A}} \\
\hline 500uA .. .. & ¢4.10 & & & \\
\hline \[
500.50 \mathrm{~A} A
\] & E4.35
\(\mathbf{4} .30\) & & & \\
\hline 1 mA .. .. .. & ¢3.95 & & & \\
\hline 1ADC" \({ }^{\text {I }}\)..... & ¢3.95 & \multicolumn{3}{|c|}{\multirow[b]{2}{*}{2}} \\
\hline 5 A DC & ¢3.95 & & & \\
\hline \(20 \vee\) DC & ¢3.95 & & & \\
\hline \(50 V ~ D C, ~\)
\(300 V\) & \(\underset{¢}{¢} \mathrm{f} .95\) & \begin{tabular}{l}
300 V AC .. \\
VU Meter
\end{tabular} & . & \\
\hline 300 V DC.. & £3.95 & VU Mater & & \\
\hline
\end{tabular}

\begin{tabular}{|c|c|c|}
\hline \(50 \mathrm{uA} . . \mathrm{.}\). .. & ¢4. 15 & \\
\hline \({ }_{2004}^{1004}\) A \(\quad .\). & ¢ \(¢ 3.75\) & ल17n? \\
\hline 500 u A & ¢ 3.50 & \\
\hline 50-0-50ua \({ }^{\text {a }}\) & f3.95 & \\
\hline 100-0-100 A A .. & £3.85 & \\
\hline 1 mA .̈ - .. & ¢ 5.50 & \\
\hline 300 V AC... .. & ¢ 63.50 & \\
\hline VU Mater .. .. & ¢4.25 & \\
\hline
\end{tabular}

\section*{MODEL ED107 EDUCATIONAL METER Size: \(100 \times 90 \times 150 \mathrm{~mm}\)
A range of high quatity}
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{3}{|l|}{A range of high quatity moving coil instruments ments and other banch applications. \(3^{\prime \prime \prime}\) mirror scale. The meter movement is easily accessible to demonstrate internal working.} & \multicolumn{2}{|l|}{\multirow[t]{2}{*}{}} \\
\hline 50 u A & & ¢7.60 & & \\
\hline \({ }^{1004}{ }^{\text {50.0. }}\) & & ¢7.05 & & \\
\hline 14.50 & & \(\underline{6.55}\) & & \\
\hline 5 S DC & & E6. 55 & & \\
\hline 5 V DC & . & E6. 55 & 300 V D & ¢6.55 \\
\hline 10 V DC & & \(\underline{66.55}\) & 500mA/5A DC & ¢7.70 \\
\hline \(15 V\) DC & . & E6.55 & \(5 \mathrm{~V} / 50 \mathrm{~V} \mathrm{DC}\) & 67.70 \\
\hline 20VOC .. & & ¢6.55 & 5V/15V OC & \\
\hline 50 VOC. & .. & £6.5 & 1A/15A OC & ¢7.70 \\
\hline
\end{tabular}

CLEAR PLASTIC MODEL 85P
50 uA A.
200uA
5000 A
500 A
50 A
\(100-0.100 \mathrm{uA}\) \(100.0-100 \mathrm{~A}\)
\(500.0-500 \mathrm{~A}\) \({ }_{1-0.1 \mathrm{~mA}}^{1} \mathrm{~A}\) 1.01 mA
5 mA.
10 mA.
\(50 \mathrm{~mA} .\).
500 A

100 mA
500 mA
1A DC
5A DC
\(15 A D C\)
\(30 A D C\)
\(10 V\) DC
\(20 V\)
\(20 V\) DC
\(50 V\) DC
\(150 V\) DC
\begin{tabular}{|c|c|}
\hline ¢4.85 & \multirow[t]{2}{*}{} \\
\hline ¢4.70 & \\
\hline E4,45 & - - - - \\
\hline E4.30
\(\mathbf{E 4 . 7 0}\) & \multirow[b]{2}{*}{x vu} \\
\hline E4.45 & \\
\hline f4.30 & \multirow[b]{2}{*}{-} \\
\hline f4.30 & \\
\hline E4.30 & \\
\hline E4.30 & \\
\hline ¢4.30 & \\
\hline E4.30 & \\
\hline c4.30 & 300 V DC .. .. \(£ 4.30\) \\
\hline E4.30 & 15 V AC .. .- \(£ 4.35\) \\
\hline E4.30 & 300 V AC .. .-. \(\quad \mathbf{¢ 4 . 3 5}\) \\
\hline ¢4.30 & S Meter ImA .. \(£ 4.30\) \\
\hline ¢4.30 & VU Meter .- .- \(¢ 5.00\) \\
\hline 14.35 & 1AAC .. .. * \(£ .30\) \\
\hline ¢4.30 & 5A AC ..- .. : ¢4.30 \\
\hline £4.30 & 10A AC .. .. * E4.30 \\
\hline ¢4.30 & 20A AC .. .. * £4.30 \\
\hline ¢4.30 & 30A AC .. .. * £4.30 \\
\hline
\end{tabular}
\(240^{\circ}\) Wide Angle 1 mA METERS MW 1-6 \(60 \times 60 \mathrm{~mm}\)
£3.97 P\&P 15p MW 1-8 \(80 \times 80 \mathrm{~mm}\) £4.97 P\& 15:


\section*{\}}


CLEAR PLAST
Size: \(42 \times 42 \mathrm{~mm}\)


2a

\section*{50
10
50}


BAKELITE MODEL S80 Enlarged Window
Size \(80 \times 80 \mathrm{~mm}\)
10
50
50
100
1 m
1 A
5 A
20
50
30
30

\section*{CLEAR PLASTIC MODEL 52P}

Size: \(60 \times 60 \mathrm{~mm}\)
\begin{tabular}{|c|c|c|c|}
\hline 50 Ma & ¢3.85 & \multicolumn{2}{|l|}{\multirow[b]{2}{*}{- \(-\cdots \cdots\)}} \\
\hline 100uA & \(¢ 3.30\) & & \\
\hline 5000 A . & ¢2.90 & - & \\
\hline \begin{tabular}{l}
\(50-0.50 \mathrm{uA}\) \\
\(100-0.100 \mathrm{~A}\)..
\end{tabular} & ¢ 83.35 & \%A & \\
\hline 1 mA & £2.75 & & \\
\hline 5 mA & ¢2.75 & & \\
\hline 10 mA ., .. & ¢2.75 & & \\
\hline 50 mA .. .. & ¢2.75 & & \\
\hline 100 mA & £2.75 & & \\
\hline 500 mA & £2.75 & & \\
\hline \(1 A D C\) & & & \\
\hline 5A DC \({ }^{\text {10V }}\).. & ¢2.75
\(\mathbf{6 2 . 7 5}\) & SMater 1 mA
VU Meter .. & E2.85 \\
\hline 20 VDC & £2.75 & 1AAC & - 12 \\
\hline \(50 \vee \mathrm{DC}\) & ¢2.75 & \(5 A A C\) & - \(£ 2.75\) \\
\hline 300V DC .. & £2.75 & 10A AC & - \(¢ 2.75\) \\
\hline 15 V AC & ¢2.85 & 20A AC .. & - 62.75 \\
\hline 300V AC .. .. & ¢2.85 & 30A AC. & - 12.75 \\
\hline
\end{tabular}
\begin{tabular}{|c|c|c|}
\hline BAKELITE N & MODEL 65 & Siza: \(80 \times 80 \mathrm{~mm}\) \\
\hline 25uA & E5.05 & \\
\hline 50uA .. .. .. & .. \(\quad 53.90\) & \\
\hline 100uA & .. \(\quad\) E 3.00 & \\
\hline 50-0.50uä \(\because\) & .. & \\
\hline 100-0.100w A.. & .. \(\quad\) E3.30 & \\
\hline 500-0-500uA .. & .. \(\quad\) E2.85 & \\
\hline 1mA .. .. .. & ¢2.85 & \\
\hline \({ }^{1.0 .1 m A}\) & E2.85 & \(\Delta\) \\
\hline 5 mA A .. .. & ¢2.85 & \\
\hline 10 mA .. & \(\underline{2} 2.85\) & \\
\hline  & .. \(\quad\) E2.85 & 50V AC .. ..: \(\mathbf{5 0 2 . 8 5}\) \\
\hline 500mA .. ... & £2.85 & 150 V AC .. .. ¢2.85 \\
\hline 1 ADC & C2.85 & 300 V AC .. .. - \(£ 2.85\) \\
\hline 2A OC & ¢2.85 & 500 V AC .. .. \(£ 2.85\) \\
\hline \(5 A D C\) & ¢2.85 & VUMeter .. .. . \(£ 4.00\) \\
\hline 15A OC .. .. & .. \(\quad \mathbf{5 2 . 8 5}\) &  \\
\hline 30A DC .. & ¢2.85 &  \\
\hline 5VDC \(50 . .\). & ¢2.85 &  \\
\hline 10V DC & ¢2.85 & 30A AC .. .. * ¢2.85 \\
\hline \(20 \vee\) DC & 22.85 & \(50 \mathrm{~A} A C \quad . \quad\).. \(£ 2.85\) \\
\hline SOV DC .. .. & ¢2.85 & \(500 \mathrm{~mA} \mathrm{AC} \mathrm{.}. \mathrm{¢2.85}\) \\
\hline 150V OC.. & ¢2.85 & \(50 \mathrm{mV} \mathrm{DC} \quad . . \quad \$ 3.20\) \\
\hline 300V DC .. .. & .. \(£ 2.85\) & \(100 \mathrm{mV} \mathrm{DC} \mathrm{.}.{ }^{\text {e }} 3.20\) \\
\hline
\end{tabular}

AUTO TRANSFORMERS
\(0 / 115 / 250 \mathrm{~V}\) Siep or step down. Fully shrouded E2.75 P\&P 18p 1000 WATT 99.50 PQP 38p E3.50 P\&P 18p 1500 WATT £12.50 P\&P 43p 2250 WATT £20.95 P\&P 50p 5000 WATT £44.95 P\&P £1

\section*{ALL PRICES EXCLUDE VAT Also see following pajes}


SWR METER Model SWR3
Handy SWR meter for
transmifter antenna align.
ment, with buitt-in field
strength matar. 5trength mater. Accuracy
\(5 \%\), Impedence \(52^{\circ}\) Indicator \(100 \mathrm{wA} \mathrm{DC.Full}\) scale 5 section collapsible
antenna. Size \(145 \times 50 \times\) 60 mm
OUR PRICE \(£ 4.25\)


MOOEL TE15
GRID OIP METER
Transistorised, Oper
atds as Grid Dip.
Oscillator Oscillator, Absorb. Oscillating Detector. Frequency range
\(480 \mathrm{kHz}-280 \mathrm{M} \mathrm{Hz}_{z}\) in six coils. 500 uA
meter. 9 V battery
operation. Size:


OUR PRICE £ 15.00
POWER RHEOSTATS
Angh quality cerami-
construction. Wind. ings embedded in
Heavy duty brush
wiper. Continuous
rating. Wide range
available ex-stock
Single hole fixing. \%" dta
Bulk quantities available.
25 WATT 10/25/50/100/250/500/ 1000 Ohms \(\quad\) E1.15 P\&P 10 p 50 WATT 10/25/50/100/250/500/
£1.62 P\&P 10p 100 WATT 1/5/10/25/50/100/250/ 500/1000/2500 Ohm
£2.34 P\& \({ }_{8}\) 15p
SH628 STEREO HEADPHDNES
 \(20,000 \mathrm{~Hz}\). Complete plug.
OUR PRICE \(\mathbf{£ 1 . 8 7}\)
LHO2S STEREO HEADPHONES Light werght headphones with padde ear pieces. \(4 / 16\). \(20-20,000 \mathrm{~Hz}\). Complote with OUR PRICE
 \(£ 1.97\)
TE1018 Deluxe Mono High Impedence Headse Sensitive magnetuc heads. with solt ear pads. I mpedence 2,60
ofms 1600 ohms 0 Cl Frequency res
\(200-4,000 \mathrm{~Hz}\) OUR PRICE E 225

OHO2S STEREO HEADPHONES and excell and excellen combined. Adjust mpedence 8 ah. impedence 8 ,
\(20-12,000 \mathrm{~Hz}\). Complete with OUR PRICE £2.25 P\&P 30p

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Moving coil. I
for language teaching. communi-
Headphone impedence 16 chme. Mac. OUR PRICE \(£ 5.95\) P\&P30p

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\section*{SUPERVALUE top gualiti TRIO communications equipment}
tRIO JR599 RECEIVER


Nine wove bazids covering 1.8-29.7
MHz \(144-146 \mathrm{MHz}\) and 10 MHz WWV
SCB SSB, CW, AM and FM. AF output is more
control. BFO. Variable \(A F\). Squalch RF
controls. 4-16 ohm output and jack for phones. Power requirement \(100 /\) \(140 \times 310 \mathrm{~mm}\).
DHPPríce 5132.51 CARA. TRIO 9R590S RECEIVER


Four bands covering 550 kHz to 30 soread on \(10,15,20.40\) and 80 mtrs.
8 . valve plus 7 diode circuit. 4 to 8 ohm output and phone jack. SSB-
CW, ANL variable BFO. SMeter and eparate band spread dial. IF freq. uency 445k Rz, audio output \(1 / 2\) watt. \(15 / 250 \mathrm{~V}\) AC, with instructions.
Our Price £42.50 paAm
TRIO JR310 SSB RECEIVER


Covers 3.5, 7. 14. 21 28.28 .5 and SSB. AM and CW. AF output more than IW. Crystal controlled BFO for SSB. SMeter, ANL etc. AC \(110 / 120-\)
\(220 / 240 \mathrm{~V}\) Size: \(330 \times 179 \times 310 \mathrm{~mm}\). OUR PRICE \(£ 75.00 \quad\) Carr. paid TRIO TR2200 TRANSCEIVER Ised portable
\(V H F\) transceive Will transmit and recerve on Stx channels botw
\(144-146 \mathrm{MHz}\). 1 watt trans. internal or external supply.
Built in charg Built-In charger
for \(\mathrm{N}_{1}\)-Cad cells
 Power/volume switch. squeich control, channel sol
ector mic. socket ector. mic. socket, Barphone/external
speaker sockat. Complate with microspeaker socket. Complete with micro-
phone and \(144.48,144.728145 .32\) crystals. Size: \(134 \times 58 \times 180 \mathrm{~mm}\).
OUR PRICE \(£ 79.50 \quad\) Carr.paid UNR30 RECEIVER


Four bonct coverine \(550 \mathrm{KHz} / 30 \mathrm{MHz}\) BFO. Bult-1n spataer. Operates from
\(220 / 240 \mathrm{AC}\). Brand naw with instructions. OUR PRICE £15.75

BELTEK W5400 CAR


Solid state mobile transicpiver tor 12 volt DC neg. Trangmits and receives
on any 12 of 28 channals between 144 and 146 MHz . Power output 10 W
and 1 W switcheble Controls: 1 . and 1W switcheble. Controls: On/ot 1 volume, squalch and channel select-
or. Internal \(3^{\text {a }}\), speaker. Complete or. Internal \({ }^{3}\) " speaker. Complete
with dynamic mic. PTT switch, three sets of crystals for \(144.48,144.6\) and 145 MHz , mounting brack at and ins.
tructions. Size: \(150 \times 70 \times 22 \mathrm{~mm}\). OUR PRICE \(£ 75.00 \quad\) P\&P 50p

\section*{LAFAYETTE HA600 Receiver}


Generat cover ape \(150-400 \mathrm{mz} .550\) anical filters. Prot front end 2 machanical filters, product detec for. Noise
limiter, variable BFO, \(S\) Mefer. Bandspread and RF gain. 220/240V AC or OUR PRICE \(£ 50.00 \quad\) P\&P 50 p BV05 Vernier TUNING DIAL App. \(7: 1\) ratio planetary
dirive vernier dial. Lo
scale \(0-180\) dienges
Blank scales \(1-5\) Dial size \(128 \times 76 \mathrm{~mm}\). Ovorall ire \(190 \times 117 \times 41 \mathrm{~mm}\). deep uncluding OUR PRICE \(£ 1.62\) P\&P 15


нISH पuality CONSTRUCTION KITS WE ARE APPOINTED
STOCKISTS ALL BRANCHES

All kits are complate with compre honsive easy to follow instructions and covarad by dull guarantee
Post and Packing 15p perk AF 20 Mono amplifier. AF 30 Mono pre amplifie AT5 Auto light control AT30 Photo cell switch unit AT50 400W Light dimmer/ ATS5 1,300W Light control/
 AT60 1 channel light controi AT65 3 channel light contro
HF61 MW transistor radio HF65 FM Transmitter.... HF75 FM transistor rad
HF310 FM tuner unit. HF325 Deluxe FM tuner unit
HF330 FM HF395 Aerial amplifier for GP310 SM bands I, II, \& ill. GP310 Stereo pre-amplifier. GT10 PSU \(100 \mathrm{~mA} / 9-12 . . . . .\). NT300 Professional stabilised
PSU. \(2 \times 30 \mathrm{~V}, 2.2 \mathrm{~A}\).......... PSU. \(2 \times 30 \mathrm{~V}, 2.2 \mathrm{~A} . . . . .\).
NT 305 Transistor converrer \(7.5 V\) or \(9 V\) DC........... NT315PSU 240V AC to 4.5AE1 100 mW output stag AE2 Pre-amplifier...... AE3 Diode receiver.
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'Amateur Electronits
The professional boak for the amateur -covers the subject from basic principles to advanced electronic technia-
ues. Complete with circuit board for ues. Complete with circuit board for
making ten of the kits listed above. OUR PRICE E3.30 [NoVAT P \& P 25 p plus VAT NOVAT

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EXPORT OROER
Philco Ford 5245 Mains Tape Recorder
2 track BSR deck with push button
controis for easy operation. Tape counter and volume control
Complete with hand mieropho direct recording lead, \(1,200^{\circ}\) reel of tape and upere reel
\(200 / 250 \vee \mathrm{AC}\) Fully OUR PRICE F17.50 anteed. OUR PRICE \(£ 17.50 \quad\) P\&P 75
Philco Ford 5248 Mains Tape Philco Ford 5248 Mains Tape Recorder. Four track,
two speeds
Piano key type
controls, tape
counter, record
ing level meter.
volume and tone controls et
Complete with
hand, microphone leed tor direct rec-
ording 1,800 of tape and spare reel \(200 / 250 \mathrm{~V}\) AC. Fully guaranteed. OUR PRICE \(\mathbf{2} 27.50 \quad\) P\&P 75 KE630 3 Station INTERCOM


Mester and rwo sub-stations Can be used on desk or wall mounted. Comp. OUR PRICE \(£ 5.25\)


OUR PRICE \(£ 24.95\) PIP tree ALSO AVALLABLE READV BUILT OUR PRICE \(£ 27.20\) P\&P 25p Also available-
SINCLAIR EXECUTIVE OUR PRICE £35.45 P\&P 25p SINCLAIR EXECUTIVE with MEMORY OUR PRICE £44.50 P\&8P25p SPECIAL PURCHASE!! BSR 8 Track Player Chassis Farmous 日S艮
Grack chassis as used in
moded TOBS
complete complete wi
silwer a nd
black escut-
cheon resdy
cheon, resady
to fit into
to fit in to
cabinet.
Output. 125 mv


OUR PRICE \(\mathbf{f 8} 95\)
P\&P50p
MP7 MIXER-PREAMPLIFIER
5 Microphone individual gain

facilitioy. Bsttery operated. SIze: 235 \(\times 127 \times 76 \mathrm{~mm}\). Inputs: Mics. \(3 \times 3 \mathrm{mV}\)
\(50 \mathrm{k} ; 2 \times 3 \mathrm{mV} 600\) ohms. 4 mV 50 k ; Phono Corsmic 100 mV 1 Meg. Output 250 mV 100k. OUR PRICE £8.97

NEW SINCLAIR PROJECT 80 HI-FI MOOULES \& PACKAGES

 Active Filter Unit... PR 55 Power Supply. p26 Power Supply.
\(2 \times 240 /\) Stereo \(80 / \mathrm{j}\) PZ \(2 \times 240 /\) Stereo \(80 / P Z 6\). Transformer for PZB..............

\section*{FM TUNER CHASSIS} 6 transistor
high quality high quality
unly. Size
andy \(153 x\) only 153 x
\(101 \times 63 \mathrm{~mm}\) 3 IF stages
Double tuned
 iscriminator Operates on \(9 V\) feed mosst amplifiers. \(108 M H z\). Ready built, ready for use.
Fantastic value for moner \(\begin{array}{ll}\text { OUR PRICE } £ 5.95 & \text { P\&P 20p }\end{array}\) A1018 FM TUNER Specification
as above FM chassis but complete in
cabinet with cabinet with
on/aff switch Size: \(184 \times\)


OUR PRICE \(£ 9.65\) P\&P 30p Stereo multiplex adapter \(\mathbf{£ 4 . 9 7}\)

OT55B DIGITAL CLOCK
MECHANISM
24 hour
alarm
setting
s.
setting
with but
in buzze
On/off
alarm 'sleep' switch. Huminated rot any dial with hours, mimutes and ssec-
onds. Autometically turns off radio, TV, light atc. and with auto-switching will turn on apain whan required. 240 V AC operation. Switch rating
OUR PRICE 55.95 P\&P 30p
BINATDNE OS100 Digital



INCORPORATING LASKYS RADIO AND G. W. SMITH \& CO. (RADIO).

AUDIOTRONIC AMR9000 Global AM/FM Portable Radio 10 wewa bands
coverin
AM:


PSB1: \(30-50\)
108 MHz . Air: \(108-136 \mathrm{MHz}\). Feature time 2 one map and timing dial. Large
clear scale. Tplascopic aerial and buitt clear scale. Telascopic aerial and buit
in asrial. AFC on \(F M .6^{\prime} \times 4\) " spasker and personal earpioce. Battery/mains
operation. Size: \(345 \times 133 \times 305 \mathrm{~mm}\). OUR PRICE £36.00

AUDIOTRONIC DOLBY B Noise Reduction Units
Reduce tape hiss by 3 dB at 600 Hz ,
6 dB at 1200 Hz and 10 dB for all freauencies above 3000 Hz . Size: \(460 \times\) \(203 \times 79 \mathrm{~mm} A G 220 / 240 \mathrm{~V}\)


PROCESS 4
For use with semi-professional tape \(-20 \mathrm{kHz} \pm 2 \mathrm{~dB}\). \(\mathrm{S} / \mathrm{N}\) better than 70 dB . Full source/tape monitoring: Switchtest tape.
OUR PRICE E50.00 P\&P 50p PROCESS 2
For use with cassette and tape recorders. Frequancy response \(30 \mathrm{~Hz}-20 \mathrm{kHz}\) Switchabe muttiplex filter. 2 Dolby calibration meters. Off tape monitor-
ing. S/N better than 70 dB . Supplied complete with test tape or cassette.
OUR PRICE \(£ 34.50 \quad\) P\&P 50p

CT5050 Cassette Recorder


Instant recording and playing. Plano key controls. Au tomatic levei control. ate control nicrophone, carrying case and shouluer strap. OUR PRICE E8.50 P\&P 50p AUDIOTRONIC ACD660 Stereo Cassette Deck


A beautifully styled tour track sterea deck with an outstanding specification offered at a remarkably low price. ing switchable noise feturers includ. ing switchable noise filter normal/
chrome tape selector, twin Vú meter stider record/playback level controls. front panel headphone socket, recording indicator lamp, phono/Din line input sockets, 3.5 mm microphone in
sockets etc. etc. Frequency responsa: sockets etc. etc. Frequency resporse:
\(40-12 \mathrm{kHz} \quad 140-16 \mathrm{kHz} \mathrm{CrO} 1) \mathrm{S} / \mathrm{N}:\) -45 dB . Crosstalk 45 dB . Separation: -35 dB . Noisa limiter -6d8 at 10 kHz . OUR PRICE E39.50

P\&P 50p


SKYFON 100 mW
OUR PRICE \(\mathbf{~} 24.95\) per pair P302 Two Channed 300 mW OUR PRICE \(£ 52.50\) per pair P1003 Three Channel 1 Watt
OUR PRICE E 71.25 per pair P\&P 50p per parr \(\frac{\text { NB. Licence requred for use in UK }}{1021 \text { Stereo Listening Station }}\) For balancing
 switehing. Two
gain controls. spaakers un oft sude
switch, stereo headphone socket. OUR PRICE £2.25 P\&P 15p AUDIOTRONIC STEREO HEADPHONES LSH2O Individual
volume controls.
sitite Stereo/mono
switch. 8 ohms. switch. 8 ohms.
\(40-19.000 \mathrm{~Hz}\). OUR PRICE £3.75 P\&P 30p LSH30 Open back type. Individual
tone and volume tone and volume
controls. 8 ohms.
\(30-20,000 \mathrm{~Hz}\) OUR PRICE \(\mathbf{£ 5 . 9 5}\) P\&P 30p LSH40 Two way
speaker system speaker sys
individual volume control
8 ohms.
\(20-2000 \mathrm{~Hz}\) OUR PRICE £7.50 P\&P 30p LSH60 \(3^{\prime \prime}\) speake
units. 8 ohms. Complere with cdse.
\(20-20,000 \mathrm{~Hz}\). OUR PRICE £8.95 P\&P 30p L. SH50 E1 powered energiser. and control unit,
with headphone with headphone
speaker selector \(4-32\) ohms
\(20-20\). \(20-24,000 \mathrm{~Hz}_{2}\) OUR PRICE £16.95 P\&P 30p LQH400 4 chann dynamic hasd. peice has four drive units.
4- \(\mathbf{3 2}\) ohms \(4-32\) ohms
\(20-20,000 \mathrm{~Hz}\) OUR PRICE E9.95



LINEAR 202 STEREO AMPLIFIER


20 Watts RMS Ster eo Amplifier with
controls for volume, balarice, bass and
 and scratch filters. Uses 20 Silicon
tranststors. Inputs for magnetic and transtsiors. Inputs for magnetic and
crystal cervidge, Aux, and tape. OUR PRICE E22.50 P\&P 50p
SPECIAL BARGAIN!! PHONIC 10 Two Way Speakers Matched par
of compact
of compaci
book-shelf

a 2 high
a
frequency
frequency un
\(\& \sigma^{\prime}\) woofer.
\(\& 6\) wooter.
8 ohms imped.
ence. Size: 348


OUR PRICE 88.85 Par P\&P 500 LEAK TRUSPEED MK3 RECORD DECK
ing
low
lpe
speed
syrchron
synchron
ous hysteresis
mor
motor giving
constant turntable
constan turntabe 33 and 45 rpm Cast Allum-
speed. in ium turntable. Fitted Goldring L75
arm and G800 cartidge, complete in
plinth with perspex plinth with perspex cover. Size 321 \%
\(400 \times 197 \mathrm{~mm}\) inc. cover. AC 100.130 \(400 \times 19\) mm
and \(200-250 \mathrm{~V}\)
OUR PRICE \(\mathbf{~} 35.95\) P\&P 75p
\(£ 11.35\)
f 18.15
GARRA/DA-HC/D.
2025TC/9TA
SP25 Mk \(/ \mathrm{G} 800 . . .\).
SP25 Mk \(3 / \mathrm{M} 44 \mathrm{E}\)
 AP 76/G800.
AP76/G800 AP \(76 / \mathrm{M} 444.7\)
AP76/M44E.

\section*{AP76/MS5E
AP76/M75ED}

AP \(76 / M 75 E\)

\section*{GL72/G800}


GL75/G800
GL78/G800

\section*{G101/P/C/G}

A105/GP2
GA 105/GP200...
GA207
RANK DOMUS
BD2000......
BA4000.....
BA4000...
WHARFEDALE
Linton W30 Teak..
Linton W30 White.

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\section*{ESSEX}

86 SOUTH ST. ROMFORO 70-20218
\begin{tabular}{|c|c|}
\hline \multicolumn{2}{|l|}{SURREY} \\
\hline \multicolumn{2}{|l|}{1046 WHITGIFT CENTRE, CROYDON} \\
\hline & 01-681 3027 \\
\hline 27 EDEN ST. KINGSTON & 01.5467845 \\
\hline 32 HILLST. RICHMONO & 01.9481441 \\
\hline \multicolumn{2}{|l|}{KENT} \\
\hline \multicolumn{2}{|l|}{S3/57 CAMDEN RD., TUNSRIOGE WELLS \(0892-23242\)} \\
\hline \multicolumn{2}{|l|}{LEICESTERSMIRE} \\
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Frequency
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20 kHz
5 mV ( \(1 \mathrm{kHz} 5 \mathrm{~cm} / \mathrm{sec}\) ). 0.3 m, conical diamond stylus. Channel separation
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\section*{RAYMER ELECTRONICS}

\section*{RE 500}

Advanced integrated Hi-F| Amplifier Only requires connection to mains speaker and input.
Specificatlons losd imp 4-16 ohm
Output 5 watts
Frequency response 20 Hz to \(35 \mathrm{kHz} \pm 3 \mathrm{~dB}\) Total distortion less than 1\% I/P Imp 100 k ohm Nom. \(28 \cdot 30\). Post free.


\section*{COMPLETE QUADRAPHONIC AMPLIFIER KIT: RE 4K}

A Hi-Fi kit comprising CBS SQ decoder. preamplifier and four channel main amplifier. All modules bullt and tested
Specifications
Similar to the Quadrix system (see below) but whith four channet main amplifier giving 40 watts total peak
power
This is an opportunity to experience quadraphonic sound al the very minimum of cost

\section*{RE QUADRIX}

This is a high quality CBS specitlcation SQ decoder with Hi-Fi stereo amplifier Designed to allow the conversion of any stereo system to full quadraphonic performance with minimum cost. Extraordinary new presence and reality of sound More than 20 Watts total peak power

\section*{Specifications}

Decoder
\(1 / P\) impedance OP impedance Nominal \(1 / P\) \(\begin{array}{ll}\text { Nominal VP } & 100 \mathrm{mV} \text { RMS } \\ \text { Front separation } & <60 \mathrm{~dB}\end{array}\) Rear separation <200

\section*{\(40 \mathrm{k} \Omega\)} 300 n \(<20 \mathrm{~dB}\)

\section*{mplifier}

Power O/P 5W RMS per channel (simultaneously) Distortion at full OP less than \(1 \%\)
Sensitivity for full O/P 100 mV
Active tone control Bass \(\pm 10 \mathrm{bB}\). Treble \(\pm 12 \mathrm{~dB}\) at \(100 \mathrm{~Hz}-10 \mathrm{kHz}\) Load impedance \(8-16\) ohms

\section*{MICROCHIP I.C.}

5 watte RMS
A general purpose Audlo c. for the constructor capabie of 5 watt RMS OP at less than \(1 \%\) total distortion torms an ideal basis for an excellent Hi-Fi system.
Free Data I.C 51.50
Data. Printed Circuit Board. 750
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ALL PRODUCTS ARE GUARANTEED AGAINST FAILURE FOR A PERIOD OF 12 MONTHS AND WILL BE SERVICED FREE OF CHARGE IF PURCHASED DIRECT AND REFUNDED
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RAYMER ELECTRONICS LTD.
Stanley Road, Cambrldge CBS ALB, England
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\hline \$STILL AT \(£ 42 \cdot 50\) \\
Solartron CD 7 IIS.2 Double \\
Beam Oscilloscope D.C.- 9 \\
Me/s; 3 MV/cm; trigger delay: \\
crystal calibrator; tin flat faced \\
tube. In good working con- \\
dition. Carr. fl. 50 . \\
\hline
\end{tabular}

NEW WIDE RANGE WOBBULATOR 5 MHz to 150 MHz up to 15 MHz sweep width. Only 3 controls, preset RF level, sweep width and
frequency. Ideal for 10.7 or \(T V\) If alignment, filters, receivers. Can be used with any general purpose scope. Full instructions supplied. Connect 6.3 V A.C. and use within minutes of receiving. All this for ONLY 65.75 p. P. \& P. 25p. (Not cased, not calibrated.)

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SINE and SQUARE GENERA. TOR. Four ranges. Independent amplitude controls, thermistor stabilised, Ready to use, 9V supply required. 27.65 each. P. \& P. 25p (Not cased, not calibrated.) GRATICULES. \(12 \mathrm{~cm} \times 14 \mathrm{~cm}\) high quality plastic 15p each. P. \& P. Sp. \(12^{\prime \prime}\) Long Persistence Crt, full spec Price \(\mathcal{E}-50\) to include V.A.T. \& carr £I WORTH OF "UFS". Six brand new capacitors all between
15 V and 100 V . Total capacitance 15 V and 100 V . Total capacitance not less than \(7,000 \mathrm{mF}\). \& P 45 p PHOTOMULTIPLIER type 931 A \& 3.25 each. P. \& P. 25p

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ew-50p. P. \& R. 17p
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PSYCHEDELIC LIGHTING UNIT IN KITFORM Make this fascinating three-channel unit from a kit which contains all its drive from the speaker terminals of a record player, tape recorder or portable radio. Will drive a total of 2 kW of coloured lamos at 240 voles. Supplied complete with PVC covered steel cabinet with holes ready punched for controls and cable outlets. Master control included Coloured lamps not supplied
Price per kit E1I.95. 240 volt coloured lamps, each 30p.

\section*{METAL CABINETS}

These attractive steel cabiners are PVC covered in a range of colours and offer an economically priced unit for the home constructor. The chassis, Which has a white satin pasimish, provides an anel. The cabinet is supplied complere with stick on feet
effort will be made tosupplyaselected Grey, Black. White and Blue. Every if possible.
}
 to your own requirements. Please send S.A.E. for quotation.

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All priced at \(£ 1.65\)
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\begin{gathered}
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\end{gathered}
\] & 0.15 \\
\hline 21 \(\frac{1}{2}\) & \(x \operatorname{lin}\) & 7p & 7 p \\
\hline 2 & \(\times\) 3数 & 27p & 20p \\
\hline \(2 \frac{1}{1}\) & \(\times \operatorname{Sin}\) & 29p & 29p \\
\hline 3 & \(\times 34\) in & 29p & 29p \\
\hline & \(\times \operatorname{Sin}\) & \(31 p\) & 32p \\
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SPECIAL SPRING FEATURES!

\section*{APRIL ISSUE}

Sound to Light Display Control Box
Highlight your parties and discos. Simplified system - easy to make, hours of fun.

Windscreen Wiper Delay
Worn-out wiper blades? Build this easy-to-use wiper control unit.
P.W. CRATA--(Cassette Recorder and Tuner Amplifier). Part 2 of this excellent new design. For more details, see March issue of P.W.

Impedance Matching
Buying hi-fi or building your own? This article irons out the bugs
SPECIALTOOL-KIT OFFER--FANTASTIC
VALUE... EXCELLENT QUALITY.
Superb kit of tools specially available to P.W. readers for constructional projects.

\section*{COMING IN MAY ISSUE}

Exclusive Supplement. Basic techniques in radio and electronics. Fully illustrated.
Oscilloscope Trace Doubler Getting the best out of oscilloscopes. Converting single beam to double beam. Double beam converted to show four waveforms simultaneously.
Distortion in Audio Amplifiers Top author sorts out a misunderstood subject and shows examples with oscillograms. Car Cassette Power Supply and Audio Booster
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\section*{APRIL ISSUE OUT MARCH 1 \\ ORDER NOW 20p}

The magazine with seven constructional articles every month!

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}

\section*{WINDSCREEN WIPER} DELAY KIT

Complete kit of parts including drilled and tinned fibreglass printed circuit board.
The unit provides intermittent wiper operations with a delay up to approximateiy 60 seconds which can be varied to optimise visibility in conditions of light rain, spray, etc. The original wiper switch overrides when normal operation is required.
83-60 including V.A.T. \& P. \& P. Cash with order.
Additional relay available for foreign cars \(£ 1 \cdot 10\) including V.A.T.
ESSEX X-RAY CO. LTD.
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AURORA
Multichannel Sound Controlled Light (PE Apr./Aug. CIB.42; 4 ch. El0.37. Reg. PSU ©3.65. PCB4 ch . control ( 47 in \(\times 101 \mathrm{in}\) ) Mk. 2 also holds



BIOLOGICAL AMPLIFIER
 ©3.20. O/P, Stages (S/c's, Rs, Cs Pots and Sw's as reqd.). Alphaphone 53p, Cardiophone 66p, Freq. Meter Cl. 60 , Vis-Feedback 55p. Audio Amp PC7 avail. to order \(\mathbf{6 5} 20\).

ELECTRONIC PIANO (PE Sep. 72/Jan. 73). Details in lists.

AMPLIFIER
(PE Nov. \(70 / \mathrm{Mar}\). 71 ) Stereo Sets \& PCBs. Pre-amp-Sc's El-80. Rs, Cs Pot and Sws-with tW MORs fll.32-with WW or tWCFRs, 60.60 PCB Cs, Pors lotin) for sets with MO and \(t W\) Rs, also holds rotary or slider pots and Maka \(^{2}\) Sw's, ©1-90. Main Amp-Rs. Cs, Pots, E5.39. PCB (3tin \(\times 5 \mathrm{in}\) ), El.15. PSU-Rs Cs, Pot, 63.97, PCB ( \(2 \mathrm{in} \times 4 \mathrm{in}\) ), 65p.
GEMINI STEREO TUNER (PE Apr./Jun. 72). PCB as publ. MICROPHONE MIXER (PE Apr. 69). S/c's, Rs, Cs, Pots, PCB (3ifin \(\times 4\) tin) also holds 6 rotary or 4 slider pots, \(63 \cdot 70\)
(PW Nov./Dec. 72). Preamp- \(5 / c^{8}\) 's (PW Nov./Dec. 72). Pre-amp-S/C's:
Rs, Cs, Pots, Sw-Mono, 22.50 ; Rs, Cs, Pots, Sw-Mono, \(\mathbf{~ S 2 . 5 0}\)
Stereo, 5.45 . PCB (3tin \(\times \frac{1}{2}\) in) Stereo, 85.45 . PCB (3ty \(\times\) thin)
(Stereo) also holds rotary or slider (Stereo) also holds rotary or slider
pots and Sw, \(\mathrm{Cl} \cdot 50\). Main Amppots and Sw, ©1.50. Main Amp-
S/c's, Rs, Cs, Pot-Mono, ©3.65; Smereo C7.30. P'PB' ( 2 f in \(\times 3 \mathrm{in}\) ) Mono). 60 p .
(PE Feb./JUR. 72) AND MODEL SERVO CONTROL

\section*{PHONOSONICS \\ PCB's AND KITS \\ AT \\ LOW PRICES}

(PE Mar./Apr. 73). S/c's, i.c.'s, Rs, Cs, Relay and pc-base. Pot Cores and pc-bases, Sw's, Pots, Panel Lamp-Mono, \(\neq 11.80\); Stereo, \(\in 18.70\).' PSU, E3.58. PCB-Main Circuit ( 3 it \(\times 9\) in) (Stereo) also holds relay and cores, \(\mathbf{£ 1 . 8 5 .}\) PCB.-Sub-Assembly, ( \(2 \frac{1}{2} \mathrm{in} \times 6 \frac{1}{2} \mathrm{in}\) ) (Stereo) Mk. 2 holds \(\mathrm{S}_{\mathrm{w}}, \mathrm{Rs}_{\mathrm{s}}, \mathrm{C}_{5}\), Presets and mounts on \(\mathrm{Sw}^{\prime}\) 's, 80p.
(PW Nov./Dec. 72). S/e's, Rs, Cs, Rotary Pots, T/fmr, 66.44: PCB(2in \(\times\) IIfin) also holds sliders, \(\mathbf{1} \cdot \mathbf{2 0}\), 9 in Spring Unit, \(\mathbf{E} 3.75\).

VIBRASONICGUITAR
(PW Sept. 70). Incl. Mic P/A, 2 -
Guitar P/A, Trem and Tone Controls Guitar P/A, Trem and Tone Controls, Rotary Pots, Lamp, Coupling 'T/fmr, 6.97. PSU, €3.57. PCB ( \(3 \frac{1}{2}\) in \(x\) 0łin) Mk. 2 , also holds 7 rotary or slider pots, \(\mathbf{C i} \cdot 75\).

\section*{ULTRASONIC}

TRANSMITTER-RECEIVER (PE May 72). S/c's, Rs, C5, Pot Relay, Dual PCB (2in \(\times 5 \frac{1}{2}\) ), \(\mathbf{E 3 . 9 0}\) T/ducers excluded.

\section*{ALSO}

PCBs as published (while stocks last) DIGICAL PSU (PE Aug. 72), 50p OSCILLOSCOPE P/A (PE Aug. 72), 33p TRIFFID (PE Feb. 73), 60p

DIGITRONIC (PW M out PCB (Itin \(\times 3 \frac{1}{2}\) in), 60 p .

SEMICONDUCTORS
\begin{tabular}{ll|l} 
ACl28 & 20p & ZTX 107 \\
ACI76 & 20p & ZTX \(\times 503\)
\end{tabular}
9p
14p
22p
\(18 p\)
\(13 p\)
22p
\(20 p\)
\(18 p\)
\(25 p\)
\(23 p\)
\(22 p\)
\(10 p\)
\(10 p\)
\(10 p\)
\(27 p\)
\(31 p\)
\(40 p\)
\(4 p\)
\(6 p\)
\(7 p\)
\(8 p\)
\(8 p\)
\(10 p\)
\(17 p\)

\section*{CAPACITORS}

ELECTROLYTIC

S.A.E FOR
FREE LIST

TAPE NOISE LIMITER (PE Feb. 72). S/C's, Rs, Cs, Pot, PCB ( \(1 \operatorname{tin} \times 3 i n\) ), E2.30. Req. PSU and

\section*{VERSATILE LIGHT EFFECTS} Single Channel SIT Channel Sound Controlled Light with built-in variable strobe. (PE June 72). S/c's. Rs, Cs, Pots, T/fmrs, Keyswitch, \(t 10.38\), PCB ( \(3 \frac{1}{2} \mathrm{in}\)
\(\times 7 \mathrm{tin}\) ) Mk, 2, also holds pors, 5 w,


PHOTOPRINT PROCESS (PE Pan
(PEjan./Feb, 72). For colour and B \& W -finds exposure, controls timing stabs. mains voltage. S/c's, SCR, LDR Rs, Cs, Pots, Relay, Keyswitch T/fmr, \(\mathbf{E 7} .50 \mathrm{PCB}\) ( 3 tin \(\times 5 \frac{1}{\text { in }}\) ) also holds pots, Sw, relay, ©i.20.

SOUND SYNTHESISER Details of PCBs \& Components in

RONDO (PE current series)
PHASING UNIT (PE Sept. 73) ND THERMOMETER
(PE SEPT.7 AND RAIN EFFECTS (PE OCT. 73

PROJECT O4

DETAILS IN LISTS



\section*{Maxi-Super HT 1810 and Maxi-Mini HT 1800 Solder Sippers}

Designed for use when working or re-working P.C. Boards. Permits removal of molten solder from Multi-leg components, enabling easy extraction. The solder is 'sipped' through the nozzle, and automatically ejected when the instrument is next used. A Swiss precision instrument manufactured to a high degree of accuracy.
The anti-corrosive outside casing has a knurled finish for more positive grip, and encases plated internal parts.
The Maxi-Super has been designed with a 3.5 kg . spring action recoilless plunger, whilst the Maxi-Mini with its conveniently shaped operating button, has a 2.5 kg . spring action plunger, protected by a channel guard. Both models have been designed with an easy-to-replace 'dupont' teflon screw-in nozzle.

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APLEASE FORWARD
without obligation
further details. PLEASE SUPPLY Maxi-Super HT PLEASE SUPPLY 1810 at \(£ 6.60\) Maxi-Mini HT 1800 at \(£ 4.95\)
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\section*{Sinclair Project 80 exciting}



Project 80 tuner
Stereo decoder


Project 80 Active Filter Unit (AFU)

\section*{only \(\frac{3^{\prime \prime}}{}{ }^{\prime \prime}\) deep \(\times \mathbf{2 ' ~}^{\prime \prime}\) high}

Living with hi-fitakes on new meaning with Sinclair Project 80 . The electronics of these revolutionary new modules are all contained within elegantly designed matching cases no more than three-quarters of an inch deep. They are designed for mounting on any appropriate flat surface by means of 6BA bolts extending from the rear of each module and which pass through suitably drilled holes. Connections are taken away out of sight in a similar manner. The possibilities opened up by Project 80 are endless - superb hi-fi systems can be installed in ways hitherto only dreamed about and never before made practical. No more cutting out and shaping to put modules in position. A few holes drilled with the aid of templates supplied and the job is done. Now you need never again be faced with problems of keeping the hi-fi from clashing with carefully thought-out furnishing schemes. (That will surely please wives!) Slider controls have been introduced in place of knobs and all modules in the range incorporate new up-dated circuitry with emphasis on performance standards and built-in protection against overload and shorting. The aim was to re-think modular construction completely - to make it infinitely more versatile, even simpler and more reliable - the result - Project 80 - another triumph for Sinclair, and the most exciting construction modules ever.

\section*{the slimmest, most elegant hi•fi modules ever made}
\begin{tabular}{|c|c|c|}
\hline System & The Units to use & Units cost \\
\hline Simple battery record player & 2.40 & \[
\begin{aligned}
& \mathbf{£ 5 . 4 5} \\
& +54 \mathrm{p} \vee \mathrm{~A} . \mathrm{T}
\end{aligned}
\] \\
\hline Mains powered record player & Z.40, PZ.5 & \[
\begin{aligned}
& \mathbf{£ 1 0 . 4 3} \\
& +£ 1.04 \vee . A . T
\end{aligned}
\] \\
\hline 30W. RMS continuous sine wave stereo amp. & \[
\begin{aligned}
& 2 \times \text { Z.40s, Stereo } \\
& 80 ; \text { PZ. } 6
\end{aligned}
\] & \[
\begin{aligned}
& \mathbf{£ 3 0 . 8 3} \\
& +£ 3.08 \text { V.A.T. }
\end{aligned}
\] \\
\hline 50W ( \(8 \Omega\) ) RMS continuous sine wave de luxe stereo amp. & \[
\begin{aligned}
& 2 \times Z .60 \text { s, Stereo } \\
& 80 ; \text { PZ. } 8
\end{aligned}
\] & \[
\begin{aligned}
& £ 33.83 \\
& \\
& £ 3.38 \mathrm{~V} . \mathrm{A} . \mathrm{T} .
\end{aligned}
\] \\
\hline Indoor P.A. & Z.60, PZ.8 & \[
\begin{aligned}
& £ 14.93 \\
& +£ 1.49 \text { V.A.T. } \\
& \hline
\end{aligned}
\] \\
\hline
\end{tabular}

\footnotetext{
Project 80 FM tuner, decoder, and A.F.U. may be added as required
}


Mount Project 80 on a bookshelf. a loudspeaker. a lampshade base a false wall with two 0.16 loudspeakers... almost anywhere.

\title{
new thinking in modular hi.fi
}

Stereo 80 pre-amplifier and control unit


Each channel has its own separate tone and volume controls operated by sliders, enabling ideal environmental matching to be obtained. A virtual earth input stage forms part of the up-dated circuitry that ensures the finest possible quality from all signal sources. Generous overload margins are allowed on all inputs. Clear instructions with template are supplied. TECHNICAL SPECIFICATIONS
Size \(-260 \cdot 50 \times 20 \mathrm{~mm}\left(10 \frac{1}{4} \times 2 \times \frac{3}{4} \mathrm{~ns}\right)\)
Finish - Black with white indicators and transparent Silders
inputs - Magnetic pick-up 3mV RIAA corrected; Ceramic pick-up 300 mV
Radıo 300 mV ; Tape 30 mV
Signal/noise ratio - 60 dZ
Frequency range -20 Hz to \(15 \mathrm{KHz} \pm 1 \mathrm{~dB} .10 \mathrm{~Hz}\) to \(25 \mathrm{KHz} \leftleftarrows 3 \mathrm{~dB}\)
Power requirements - 20 to 35 votts
Outputs \(-100 \mathrm{mV}+A B\) monitoring for tape
Controls - Press button for tape radio and P.U. Sliders for volume,
bass ( -12 dB to -14 dB at 100 Hz ) treble ( +11 dB to -12 dB at 10 KHz )


\section*{Project 80 FM tuner}


Making the Project 80 F.M. tuner and decoder avallable separately gives a wider choice of systems and saves money where stereo reception may not be required. The tuner is a triumph of electronic design and assures excellent performance. The decoder gives a 40 dB channel separation with 150 mV output per channel. Both units may be used with other than Project 80 systems.
TECHNICAL SPECIFICATIONS OF TUNER
Size \(-85 \times 50 \times 20 \mathrm{~mm}\left(3 \frac{1}{2} \times 2 \times \frac{3}{4} \mathrm{n} \mathrm{ns}\right)\)
Tuning range -87.5 10 108 MHz
Detector - I.C balanced coincidence for good A. M. rejection
Onel.C. equal to 26 transistors
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4 pole ceramic filter in I.F. section
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Fig. 1 Circuit diagram of lamp inverter. Component values applicable to $8 \mathrm{~W}, 13 \mathrm{~W}$ or 20 W umits are given in


## २ IGNITION SYSTEM


'nterwiring details of the ignition system. Also shown are ur and thyristor mounting details

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Fig. 4 Transformer winding details for the 6V version


Fig. 4 Transformer winding details for the 12 V version

## COMPONENTS...

12V VERSION GV VERSION
Resistors

| R1 | $22 \Omega$ | $33 \Omega$ |
| :---: | :---: | :---: |
| R2 | $150 \Omega 2 W$ | $100 \Omega 1 \mathrm{~W}$ |
| R3 | $150 \Omega 2 \mathrm{~W}$ | $100 \Omega 1 \mathrm{~W}$ |
| R4 | $1 \mathrm{k} \Omega$ | $1 \mathrm{k} \Omega$ |
| R5 | $470 \Omega$ | $470 \Omega$ |
| R6 | $1 \mathrm{k} \Omega$ | $1 \mathrm{k} \Omega$ |
| R7 | $1 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ |
| R8 | $1-2 \mathrm{k} \Omega$ | $1 \mathrm{k} \Omega$ |
| R9 | $150 \Omega 2 W$ | 33^2W |
| R10 | $15 \Omega 1 \mathrm{~W}$ | $5 \Omega$ 2W |
| R11 | $100 \Omega$ | $100 \Omega$ |
| R12 | 330k $\Omega$ | $330 \mathrm{k} \Omega$ |

All $\pm 10 \% \frac{1}{2} \mathbf{W}$ carbon unless otherwise specified Capacitors

| C1 | $0.22 \mu F$ | $0.33 \mu F$ |
| :--- | :--- | :--- |
| C2 | $0.1 \mu F$ | $0.22 \mu F$ |
| C3 | $0.47 \mu F$ | $0.47 \mu F$ |
| C4 | $220 \mu F 40 V$ elect. | $220 \mu F 40 V$ elect. |
| C5 | $0.01 \mu F 1000 V$ | $0.01 \mu F 1000 V$ |
| C6 | $1 \mu F 440 V$ a.c. rated | $1 \mu F 440 V$ a.c. rated |
| C7 | $220 \mu F 40 V$ elect. | $220 \mu F 40 V$ elect. | Transistors

TR1, TR2 2TX500
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TR3 $2 T \times 500$
Ferranti
TR1, TR2 ZTX 500
Ferranti (2 off)
TR3 BFS97 or
ZTX500

TR4, TR5 2N3055 (2 off) TR4, TR5 2N3055 (2 off)
Diodes and Thyristor
D1, D2 $\mathbf{Z S 1 7 0}$ or ZS270 Ferranti (2 off)
D3 - D6 $\mathbf{Z S 1 7 8}$ or $\mathbf{Z S 2 7 8}$ Ferranti ( 4 off)
CSR1 BStB0246 Siemens

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* Automatic Circuit Protection


## COMPONENTS...

| 8 WAT'T | 13 WATT | 20 WATT |
| :---: | :---: | :---: |
| Resistors |  |  |
| R1 $33 \Omega \frac{1}{2} \mathrm{~m}$ | 4.7Siw | 48752 w |
| R2 10k $\Omega \frac{1}{2} \mathbf{w}$ | 10k $\Omega \frac{1}{2} \mathrm{w}$ | $10 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W}$ |
| Capacitors |  |  |
| C1 125 $\mathrm{HF} / 16 \mathrm{v}$ | 125 $\mu \mathrm{F} / 16 \mathrm{v}$ | 250MF/16v |
| C2 | $0.02 \mu \mathrm{~F} / 30 \mathrm{v}$ | 0.05 $\mu \mathrm{F} / 30 \mathrm{v}$ |
| C3 0.04 $\mu \mathrm{F}$ | 1 $\mu \mathrm{F} / 10 \mathrm{v}$ | 2 $\mu \mathrm{F} / 10 \mathrm{v}$ |
| Transistors |  |  |
| TR1 MJE 520 | BDY 62 | BDY 62 |
| Diodes |  |  |
| D1 IN 4001 | IN 5401 | REC 31 |
| D2 IN 4001 | IN 4001 | IN 4001 |
| Transformer (T) |  |  |
| L1 10T/26s.w.g. | 10T/22 s.w.g | 10T/22 s.w.g. |
| L2 14T/30 s.w.g. | 15T/26 s.w.g | 15T/26 s.w.g. |
| L3 105T/30s.w.g. | 55T/30 s.w.g | 55T/30 s.w.g. |
| $446 \mathrm{~T} / 30 \mathrm{s.w.g}$. | 5T/30 s.w.g. | 5T/30 s.w.g. |
| Ferroxcubes FX2240 | FX2241 | FX2241 |
| (both ferroxcube pot cores and bobbins to suit can be obtained from Gurneys Radio, 91 The Broadway, Southall, Middlesex). |  |  |
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## $P=$



Fig. 1 Circuit diagram for the Scorpio Mark 2 ignition system. The component values shown are for the 12 V version, for the 6 V version see the accompanying component list


Fig. 5 transist


Collector
(Case)


Fig. 3 Layout of the components on the pi circuit board


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    88 to $108 \mathrm{Mc} / \mathrm{M}$ Britiah made. 2 Transintort resdy alisned -requires $10.7 \mathrm{Mc} / \mathrm{a}$ 1.7. Comploto with taning gang. Connections supplied bat some technics experionce encential.Our price $£ 3.95$
    Pont 20p
    R.C.S. STABILISED POWER PACK KITS

    All parte and instructions with Zoner Diode, Printed Circuit, Bride Rectifert and Double Wound Maina Transiorme or 15 or 18 or $80 V$ d.c. at 100 mA or less \& 9.97 Post
     Detail! 8.A.E. Bise $31 \times 1 \frac{1}{1} \times 1 \mathrm{in}$. PRE-AMPLIFIER BRITISH MADE Ideal for Mike, Tape, P. U., Guitar, ete. Can be ured with
    Bettery $9-12 \mathrm{~V}$ or H.T. line $200-300 \mathrm{~V}$ d.c. operation. siee: Battery $9-12 \mathrm{~V}$ or H.T. line $200-300 \mathrm{~V}$ d.c. operation. 8ive $1 \frac{1}{2} 1 \ddagger x$ in. Response $25 \mathrm{c} / \mathrm{a}$ to 25 ko/a, 26 dB gain. Por use with Falve or tranaistor equipment.
    Full inatructions tapplied. Details 8.A.E.
    NEW TUBULAR ELECTROLYTICS CAN TYPES

    |  |  |  |  |  |  |
    | :---: | :---: | :---: | :---: | :---: | :---: |
    | $2 / 350 \mathrm{~V}$ | 14D | 250126V | 14p | $50+50 / 800 \mathrm{~V}$ | 50 p |
    | $4 / 350 \mathrm{~V}$ | 14p | $500 / 25 \mathrm{~V}$ | 20p | $60+100 / 350 \mathrm{~V}$ | 85 D |
    | $8 / 450 \mathrm{~V}$ | 18 p | 1000/25V | $35 p$ | $38+82 / 250 \mathrm{~V}$ | 0, |
    | 16/450V | 82p | 1000/60V | 470 | $32+82 / 450 \mathrm{~V}$ | 60 p |
    | 32/500V | 500 | $8+8 / 450 \mathrm{~V}$ | 22p | $350+50 / 825 \mathrm{~V}$ | 559 |
    | $25 / 25 \mathrm{Y}$ | 100 | $8+16 / 450$ | 250 | 1019 | 45 | $25 / 25 \mathrm{~V} \quad 50 \mathrm{D}(8+8 / 450 \mathrm{~V} \quad 22 \mathrm{p} \quad 350+60 / 826 \mathrm{~V} \quad \mathrm{S5p}$ $50 / 50 \mathrm{~V} \quad 10 \mathrm{D}(16+16 / 450 \mathrm{~V} \quad 40 \mathrm{p} \quad 100+50+50 / 350 \mathrm{~V} 55 \mathrm{D}$ $100 / 25 \mathrm{~V} \quad 10 \mathrm{p}, 32+82 / 350 \mathrm{~V} 40 \mathrm{p}$

    LOW VOLTAGE ELECTROLTMICS.
    $1,2,4,5,8,16,25,30,50,100,800 \mathrm{mF} 15 \mathrm{~V} 10 \mathrm{p}$
    
    $1000 \mathrm{mF} 12 \mathrm{~V} 80 \mathrm{p} ; 25 \mathrm{~V} 35 \mathrm{p} ; 50 \mathrm{~V} 47 \mathrm{p} ; 100 \mathrm{~V} 70 \mathrm{p}$.
    2000 mF 6V 25p; 25V 42p; 50 V 57 p .
    2500 mF SOV 62p; $3000 \mathrm{mF} 25 \mathrm{~V} 47 \mathrm{p} ; 50 \mathrm{~V} 65 \mathrm{p}$
    5000 mF 8V $25 \mathrm{p} ; 12 \mathrm{~V} 42 \mathrm{p} ; 25 \mathrm{~V} 75 \mathrm{p} ; 35 \mathrm{~V} 85 \mathrm{p} ; 50 \mathrm{~V}$ 05p.
    CERAMIC 1 pF to 0.01 mF , 4 p . Silver Mica 2 to 5000 pF , 4p. PAPER 350V-0.1 4p; $0.513 \mathrm{p} ; 1 \mathrm{mP} 15 \mathrm{p}$; \&mP 150 V 15p $500 \mathrm{~V}-0.001$ to $0.054 \mathrm{p} ; 0.15 \mathrm{p} ; 0.258 \mathrm{p} ; 0.4725 \mathrm{p}$.
    ILVER MICA. Close tolerance $1 \%$. $2.2-500 \mathrm{pF}$ 8p; $580-$ $2,200 \mathrm{pF} 10 \mathrm{p} ; 2,700-5,600 \mathrm{pF} 20 \mathrm{p} ; 6,800 \mathrm{pF}-0.01$, mid 30 p each TWIN GANG. " $0=0$ " $208 \mathrm{pP}+176 \mathrm{pF}$, 65 p .
    Slow motion drive $885 \mathrm{pF}+865 \mathrm{pF}$ with $85 \mathrm{pF}+25 \mathrm{pF}, 50 \mathrm{p}$;
    HORT WAVE SINGLE $10 \mathrm{pF}, 300$. $25 \mathrm{FE}, 55 \mathrm{p}$
    SHORT WAVE single GANG. Precinion Silver Pleted Gangable Tuning Condensers.
    section Couplers supplied EREE with two or mp each
    NEOM PANEL INDICATORS 250V AC/DC. Amber 20p. EESISTORS. $\frac{1}{W}, \frac{1}{3} \mathrm{~W}, 1 \mathrm{~W} .20 \% 1 \mathrm{p} ; 2 \mathrm{~W}, 5 \mathrm{p} .10 \Omega$ to 10 M HIGH STABILITY. 1 W $2 \% 10$ obms to 6 mes., 10
    Ditto $5 \%$. Preterred 7alues 10 , 5 watt, 10 watt, 15 watt 10 ohms to $100 \mathrm{~K} 10 \mathrm{peach} ; 0.5$ ohm to 8.2 ohms 10 p . TAPE O8CILLATOR COLL Valve type 35 p .
    FERRITEROD $8 \times$ in $20 \mathrm{p} ; 6 \times \sin 20 \mathrm{p} ; 3 \times 1$ in 10 p .

    ## MAINS TRANSFORMERS $\begin{gathered}\text { all posp } \\ 25 \mathrm{p} \\ \text { osch }\end{gathered}$

    Eagle MT12 12-0-12V 60 mA8.3 V 3.5 A 6.8V 1 A or 5 V QA
    
     MINIATURE $200 \mathrm{~V} 20 \mathrm{~mA}, 6.3 \mathrm{~V} 1 \mathrm{~A} 21 \times 21 \times 2 \mathrm{Bin}$ MIDGET $220 \mathrm{~V} 45 \mathrm{~mA}, 6-3 \mathrm{~V} 2 \mathrm{~A} 2 \mathrm{~s} \times 24 \times 2 \mathrm{in}$
    HEATER TRANS. 6.8V 3A
    GENERAL PURPOSE LOW VOLTAGE. TBD...... 80p
     amp, 6, 8, $10,12,18,18,20,24,80,86,40,48,6024-80$ amp. $6,8,10,12,16,18,20,24,80,86,40,48,6022-25$ 5 amp, $6,8,10,12,16,18,20,24,30,86,40,48,6028880$
    
     $6-0-6 V 500 \mathrm{~mA} 90 \mathrm{p}$ oV 1 amp 96 p 12 V 300 m.
    500 mA 85 p 12 V 750 mA 95 p .40 V 1 amp cl .75.
    AOTO TRANSFORMERS. 115V to 230 V or 230 V to 115 V 150 W 22.80; 500 W 26.25; 750W $£ 10$; 1000 W 215 CRARGER TRANBFORMERS. Inpat 200/250V. for 6 or 12V, 1 \& amp 21.50; $2 \mathrm{amp} \mathrm{sl} \cdot 80 ; 4 \mathrm{amp} 22.50$ BATTERY CHARGERS. Ready built with leads and clipa 1; amp ef: 4 smp $44 ; 5$ amp. e4.50.
    FULL WAVE BRIDGE CHARGER RECTIEIERS:
    6 or 12 V outpats. It amp $40 \mathrm{p} ; 2$ amp $55 \mathrm{p} ; 4 \mathrm{amp} 85 \mathrm{p}$.
    MAINS ISOLATING TRANSFORMER
    Primary 0-110-240V. secondary $0-240 V 8$ amps 720
    watta. Insulated terminals. Varnith impregnated. Fally
    andor onclosed in steel cate with fxing leet. $\mathrm{Cl} \mathbf{O}^{\mathrm{Carr}}$. Pamous maxe (Value eng) OUR PRice

    ## SET OF 3 MOTORS FOR

    COLLARO STUDIO TAPE DECK
    £2.50 Post 50 p
    VOLUME CONTROLS 80 hm Coax 4 pyd.
    Long epindler. Midget Size
    5 K . ohmito 2 Meg. LOG or LIN. L/S 15p. D.P. 25p. Edge 5K. S.P.Transiator 25p. BRITISH AERIALITE AERAXIAL-AIR SPAC 40 yd $81 \cdot 40 ; 60$ gd $\& 2$ FRINGE LOW LOS8 10 per
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    HELICAL POT 10K LIN. List \&3. Bargain 75p.
    DUAL CONCENTRIC POT 500K LOG AND 500K LIN D.P. switch. Inner spindle 3\}in; outer spindie 2tin 75p.
    E.M.I. $13 \frac{1}{2} \times 8$ in. SPEAKER SALEE ! Mink buninineotarit $£ 4.50$ And cromover. 8 or 8 or 15 ohm . As illuatrated. Pont 25 p With flared tweater cone and cersmic marnet. 10 watt. Beal ren. $45-60 \mathrm{c} / \mathrm{s}$. State 3 or 8 or 15 ohm. Pont 25p
    Bookshelf Cabinet ${ }^{18} \times 10 \times$ gin $\neq 5.50$
    GOODMANS $6 \frac{1}{2}$ in. HI-FI WOOFER
    8 ohm. 10W. Large ceramic magnet Special Gambric cone aurronnd. Frequency reaponse
    $30-12,000$
    Ideal P.A.
    

    Colomis Hi -Fi Enclolure syteme, etc
    Suitable Cabinet $12 \times 8 \times 884$ Suitable Tweter 82
    

    ## ELAC CONE TWEETER

    The moving coil diaphragm gives a good radiation pattern to the higher trequenciea from $1,000 \mathrm{c} / \mathrm{s}$ to $18,000 \mathrm{c} / \mathrm{s}$. $81 \mathrm{se} 33 \times$ $3 \pm \times$ 8in deep. Rating $10 \mathrm{~W}, 8$ ohm. Cromover 51.85 $\quad \mathbf{8} .90$ Post 80p.

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    8 in. WOOFER 8 ohm 12 Watt. Deep cons. Heary ceramic magnet. Bas resonance 35 cph . Frequency rapponse $30-8,000$ epr. deal base unit for $\{3 \cdot 75$ Hi-Fi system
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    8 ohm, 21 in ; 2 iin: 3 in; $5 i n$.
    ohm, $2 \frac{1}{2 n} ; 2 \operatorname{lin} ; 5 i n \times 8 i n ; 5 i n$.
    $15 \mathrm{ohm}, 8 \mathrm{in} ; 51 \mathrm{n} ; 8 \times 4 \mathrm{n} ; 2 \times 4$
    5 ohm, $2 i n ; 3 i n ; 5 \times 3 i n ; 5 i n$
    
    £ 1 each
    LOUDSPEAKERS P.M. 8 OHM8. $7 \times 4$ in 81.25 ; 8 Hin 81.50 $8 \times \operatorname{Sin} 81-60 ; 8 \mathrm{in} 51.75 ; 10 \times 6 \mathrm{in}$ £1.90; 10 in 82.00 RICHARD ALEAN 10 in diameter 4 W 22.50, 10 in diameter 5W 22.50; Post 25p 2in diameter, 6 , 2,8 , modela SPEAKER COVERING MATERIALS. Samples Large S.A.E.
     TWO-WAY 3,000 c.p.s. CROSSOVERS 8, 8 or 15 ohm $\$ 1.85$

    TWO.WAY CROSSOVER NETWORK $3,000 \mathrm{c} / \mathrm{s}$ With variable tweeter attenuator giving accurate high/low
    frenuency balance. Mounted on panel $\sin \times 4 \operatorname{in}$ with control Enob, tweator and wooler leade and input $2 \cdot 20$ Post terminals. 8uitable for 8 to 8 ohm impedance. $22_{20 p}^{20}$

    VALVE OUTPUT TRANSFORMER $25 p$ p
    MIKE TRANSFORMER MO motal $100-1$ \&1-25. SWATT MULLI RATIO. 8,8 and 15 ohme,
    50 wett …….... $18.50 \quad 100$ wett

    ## ELECTRO MAGNETIC

    PENDULUM MECHANISM1-5V d.c. operation over 800 houra continuous on 8P2 battery, fully adjustable owing and apeed. Ideal dapley, Post strobe, etc.

    ## R.C.S. RECORD PLAYER AMPLIFIER

    2 stage triode pentode valve. 8 watte output. Volume on/on and tone controls. Printed circuit
    

    COAXIAL PLUG 10p. PANEL SOCKETS 10p. LINE 18p COAXIAL PLUG 10p. PANEL SOCKETS 10p.
     JACK SOCKET Std. open-circuit 14p. closed circuit 23p; Chrome Lead Socket 45 p . Phono Pluga 6 p . Phono 8ocket 6p JACK PLUGS Std. Chrome $20 \mathrm{p} ; 3-5 \mathrm{~mm}$ Chrome 12 p DII SOCKETS Chassie 3-pin 10p; S-pin 10p DIN SOCKETS Lead 3-pin 18p; 5-pin 15p. DIN PLUGS 3-pin 18p; 5-pin 25p. VALVE HOLDERS 5D; CERAMIC 8p; CANS 5p.
    E.M.I. $\underset{\substack{\text { WOOOER } \\ \text { TWEETER KIT }}}{\text { Kit }}$ £5.75 (Available separately. Woofer 84.25 ; Tweeter \$1.90)
    Comprising a fue example of a Wooter $10 \frac{3}{2} \times 6$ !in with a massive Ceramic Magnet, 440 z Gauls 18,000 lines. Aluminiam Cone centre to improve middle and top response. Also the E.M.I. Weeter stin square has a Epecial lightweight paper cone and magnet flux tull instructions appplied.
    Impedance Standard 80 hms
    $\begin{array}{ll}\text { Maximum power } & 12 \mathrm{wath} \\ \text { Mat }\end{array}$ Useful Response 35 to $18,000 \mathrm{cps}$ Basa Reronance 45 cpa
    8UITABLE ENCLOSURE $20 \times$ 8UITABLE ENCLOSURE $20 \times 13 \times 9$ in MODERN DESIGN. TEAK WOOD FINISH
    
    $\$ 9.90$
    Pont 25p
    

    8in. or IOin. ELAC HI-FI SPEAKER
    Dusl cone plasticised roll surround. Large cersmic magnet. $50-16,000 \mathrm{cps}$. Bats resonance 55 cpu .8 ohm impedance. $10 \mathrm{in}, 12 \mathrm{watt}$.
    63.75

    TEAK VENEER HI-FI SPEAKER CABINETS
    Fluted Wood Fronts
    MODEL "A". $20 \times 13 \times \operatorname{Oin}$ For 12 in. dia. or
    Oin speazer. $\mathbf{~} 9.90 \begin{array}{r}\text { Post } \\ 25 p\end{array}$ MODEL "B". $16 \times 10 \times$ in For $18 \times 8$ in, or $\quad \$ 5.50$ Post MODEL " $B$ " 2 ditto. Triangalar Corner Vercion.
    $\begin{array}{ll}\text { MODEL "C". } & 16 \times 8 \times 6 i n . \\ \text { For } 8 \times \text { sin. } & \mathbf{4 4}\end{array}$ peaker $\quad 64.40$ Pont
    LOUDSPEAKER CABNET
    WADDING 18 in wide, 15 p it.
    BARGAIN AM TUNER. Medum Wave.
    Transistor 8 uperhet $\leqslant 4.95$ errito aerial. 9 volt. 95 BARGAN 4 CHANNEL RAN8ISTOR MONO MIXER. Add musical highlighte and zound effect to recordings. Will mix nd tuner with sepsrate control into singlo output. 9 volt bettery $\quad \leq 3.95$
    STEREO VERSION OF ABOVE $\mathbf{5 5} .95$
    BARGAIN FM TUNER.
    8-108 - $\mathrm{c} / \mathrm{six}$ sraniator.
    volt. Printed cireuit.
    alibrated glide da tuning.
    Walnut Cabinet. $\leq 12.85$
    IEE $7 \times 5 \times$ Yin. TUNER
    AARGAIN FM TUNER $\mathbf{~} 8 . \overline{85}$
    Aa above leas cabinet.
    

    BARGAIN 3 WATT AMPLIFIER. 4 Tranisto Push-Pull Ready built with volume, treble and $\quad \mathbf{~} \mathbf{4 0} \mathbf{5 0}$
    basp controls, 18 volt battery operated.

    THE "INSTANT" BULK TAPE
    ERASER \& HEAD DEMAGFETISER.
    aitable for cassettes, and sll sizes of Leaflet 8.A.E.
    €3.50 ${ }_{20}^{\text {Pont }}$

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    OFFERENG 1001 USES for overy type of heating and drying applications in the home, garage, greenhouse actory (svailable in manufacturing quantities). Approx 250 watts $8, \times$ is in. ubentos fitted with connecting wires Completely ferible providing safe Black heat. British-made for use in photocopiers and print drying equipment.
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    ONLY 40 EACH (FOUR FOR EI.50) ALL POST PAID-Discounts for quantity

    BAKER MAJOR $12^{\prime \prime}$ \&9.90
    
    
    $30-14,500 \mathrm{c} / \mathrm{s}, 12 \mathrm{in}$. Dont Free cone, woofer and tweeter cone together with a BAKER ceramic magnet assembly 14,000 gans and a total flux of 145,000 Mexwella. Baes resonance 40 c/a Rated 20 watts. NOTE: 3 or 8 or 15 ohms must be stated.

    Module kit, 30-17,000 c/s with tweeter, crossover, baffe and ingtructions. $\leq 12.50$ Post Free

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    ## "GBIG SOUND"

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    Robuatiy constructed to stand up to long periods of electronic power. Aseivl by leading groups. Ueelul response $30-13,000 \mathrm{cps}$.
    Bass Resonance 55 cps .

    GROUP "25" 12in 25 watt 3,8 or 15 ohms. 68.80 Post Free
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    12in 35 watt 69.90 Post Free
    GROUP "50" 15 in. 50 watt
    8 or 15 ohms.
    £22.00
    NEW I2in VERSION E17.60

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    ALL PURPOSE TRANSISTOR AMPLIFIERAll purpose transiatorised Ideal Ior Groups, Disco and P.A. 4inputs speech and music. 4 way mining. Ousput $8 / 15$ ohm a.c. Mains. Guaranteed. Detaila S.A.E.
    

    CALLERS ONLY! DE-LUXE 100 WATT AMPLIEIGR CHASSIS. 7 Valve veraion, 4 inputs, 10 wide range 4,8 and 15 ohm Loudgpeg, Organs, Guitars, etc.

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    Size: ?* A die, punch and Allen Screw
    

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    100 ohm Rheostat 21 in dis. Ceramic former screw terminals 1 in dia. spindle. 95 p. Post 25 p.

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    3 in Pazel Meter $50 \mu \mathrm{~A}$. Unusual seale requires re-calibration. 21-75. Post 25p.

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    ## DELUXE 4 POLE MOTOR

    ;1,400 r.p.m. reversible 42 Watt. 3 in As illustrated, $240 \mathrm{~V}_{\text {a }}$ mains.E.M.I. GRAM MOTOR

    120 V or 240 V s.c. $\mathbf{2 , 4 0 0 \mathrm { rpm } . 2 \text { -pole }}$
    120 V or 240 V s.c. $2,400 \mathrm{rpm}$.
    70 mA . 8ize $2 \mathrm{i} \times 2 \mathrm{t} \times 2 \mathrm{in}$.
    1.95

    Post 25p
    \& 1.00
    E1.00

    BAKER HR-FI SPEAKERS
    HIGH QUALITY-BRITISH MADE REGENT
    I2in. 15 watts
    An inexpenaive unit for the beginner in high fldelity and for general purposes. May be Amplifler $\mathrm{Hi}-\mathrm{Fi}$ or Teleriaio Ampiner, Hi-F or Televinion
    Bage Resonance
    lux Density 12,000 gsep
    Useful response $45-18,000 \mathrm{cps}$ 3 or 8 or 15 ohm models.
    £8.80 品品
    DE-LUXE Mk II 12in. 15 watts
    Especially designed to provide all range reproduction at an se with sng high fidelity ystem. Built-in concentric weeter cone.
    Basa Resonance
    Flux Dengity $\quad 14,000$ gang Ureful remponse 25-16,000cp 8 or 15 ohms models.
    

    ## SUPERB

    I2in. 20 watts
    A high quality loudspeaker, its remarkable low cone resonance ensures clear basis. Fitted with a special copper drive and concentric tweeter cone refulting in full range reproduction with remarkable efficiency in the upper register. Flux Density 10,500 25cpa Flux Density $\quad 10,500$ gauss 8 or 15 ohms modela.

    ## f16.50

    

    ## AUDITORIUM

    I2in. 25 watts
    A fnll range reproducer for high power, Electric Guitara, public address, multi-speaker syatems, electric organs, Ideal for $\mathrm{H}-\mathrm{Pi}$ and DiscoBase Reio
    Blasi Renonance $\quad 35 \mathrm{cps}$ $\begin{array}{ll}\text { Flox Density } & 15,000 \text { ga uss } \\ \text { Oieful response } & 25-16,000 \mathrm{cpa}\end{array}$ 8 or 15 ohms models.

    ## £ $15 \cdot 40$

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    A high wattage loodspeaker of ezceptional quality with a level response to above $8,000 \mathrm{cpa}$. Ideal for Public Aronic instruments romic instruments and the Basa
     Useful responce $\quad 20-14,000 \mathrm{cps}$ 8 or 15 ohms models.
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    Hi-Fi Enclosure Manual containing 20 plens, derigna Hi-Fi Encloaure Manual containing 20 plan
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    CUSTOMERS FREE CAR PARK
    

    ## 3/ CARRY OUT OVER <br> 40 EXPERIMENTS ON BASIC ELECTRONIC CIRCUITS \& SEE HOW THEY WORK, including :

    valve experiments, transistor experiments amplifiers, oscillators, signal tracer, photo electric circuit, computer circuit, basic radio receiver, electronic switch, simple transmitter, a.c. experiments, d.c experiments. simple counter time delay circuit. servicing procedures

    This new style course will enable anyone to really understand electronics by a modern. practical and visual method-no maths, and a minimum of theory-no previous knowledge required It will also enable anyone to understand how 10 test, service and maintain all types of electronic equipment, radio and TV receivers, etc
    

    ## INSPIRING TIMES

    F necessity be the mother of invention, we are now surely entering upon a period of prolificity in circuit creation and adaptation.

    It is the amateur designer and experimenter who so often has the knack of hitting upon the right kind of solution to a sudden problem affecting everyday life. And problems there certainly are at this present time when restrictions in the use of electricity, if not complete cuts in the supply, affect all households; and soaring prices, if no actual shortage of petrol, make every motorist that more efficiency conscious.

    This month's Ingenuity Unlimited includes a novel circuit devised by one of our readers and it is likely to be of interest to those with electrically controlled gas fired heating systems. A fore-runner, no doubt, of other crisis inspired ideas. We shall be happy, indeed consider it to be our duty, to publish other circuit ideas which may help overcome hardships or inconvenience experienced at home or on the road.

    There already exist, of course, many circuits and systems that have relevance in the present difficult times. As a service to our readers we have prepared a special Blueprint for inclusion in this issue. It gives details for constructing two extremely useful and topical projects.

    One side of this Blueprint carries details of the P.E. Scorpio Mk 2 Ignition System. The great success of the first Scorpio ignition system coupled with a continuing demand for copies of the original article made it quite clear that publication of an up-dated version would be highly welcomed; and in actual fact, could be a genuine service to the car owning public. The merits of the capacitor discharge ignition system are now well established, but they are, in any case, summarised in the article which complements the information given on the Blueprint.

    The other side of the Blueprint is devoted to a design for a lamp inverter. Such devices are not by any means novel, but since design information may not be immediately to hand at the crucial moment, we expect a number of readers will appreciate the appearance of this useful design.

    ## NEW PRICE

    We ourselves, alas, are not immune from the effects of rising prices and shortages of raw materials, and we regret having to end this month with the following domestic note.

    An apology is due to our readers for the lack of advance warning with regard to an increase in price of Practical Electronics. This increase to 25 p takes effect as from this (March) issue. Unfortunately it was not possible to make an earlier announcement in this respect.

    We trust all our readers will understand that the current high cost of materials, especially paper, makes this new price for P.E. unavoidable.
    F.E.B.

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    F. E. BENNETT

    ## Editorial

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    Probably the simplest rhythm generator is the old fashioned metronome. This device maintains the player's tempo but it cannot emphasise the down beat or create any special effects.
    Down beat enhancement can be achieved electronically but for more entertaining rhythm accompaniment it is more satisfying to tie the output of an electronic metronome to electronic percussion generators in a selected sequence or rhythm pattern such as Waltz, Cha-cha, etc.

    All commercially available switched rhythm generators have fixed rhythms which cannet be changed except by rewiring the switches. The generator to be described has no such limitations, it can produce any rhythm pattern of 16 pulses, for a total of 64,000 patterns, to be used to drive any one of eight percussion generators. Details for setting up the common rhythm patterns will be given.

    Four time signature switches select $4 / 4,3 / 4,6 / 8$ or $9 / 8$ time for the 16 available pulses and when programming a rhythm pattern the pulse order can be monitored using a Minitron indicator. Because circuit layout is not particularly critical detailed constructional information has been omitted and instead use has been made of photographs annotated where necessary, to provide rudimentary guidelines for assembly.

    ## BLOCK DIAGRAM

    Fig. 1 shows the block diagram for the Rhythm Generator. The tempo generator is an assymetrical astable multivibrator. It will start when power is applied and run at all times from that point on. The frequency is controlled by the tempo control VR1.

    The output of this generator is fed to a divide by four circuit. This expedient was used to keep the
    size of the tempo generator components to reasonable levels and insure more reliable operation of that part of the circuit. The tempo generator runs then at an effective 64 beats per measure.

    The four bit binary counter is used to generate 16 binary numbers which are decoded for the address information (storage address) and applied to IC6 for decoding and application to the Minitron indicator. IC5 decoding chip is used to supply reset pulses to the counter for 9 and 12 counts giving $9 / 8$ time and $3 / 4$ or $6 / 8$ time respectively.
    The step generator is switched to the input of the counter when the "Stop/Run" switch is placed in the "Stop" position. The divide by four output is removed at that time and the counter can be manually stepped from 0 through 15.

    ## MEMORY CHIPS

    The storage section for the programmed rhythms is made up of two SN7489, 64 bit bipolar memory chips ( $1 \mathrm{C} 7 / 1 \mathrm{l} 8$ ). Each of these contains 64 flip-flops capable of storing a binary 0 or $1(0$ volts or 4 volts respectively). The particular group of four flip-flops selected per chip (eight for both chips) is controlled by the address lines which get their information from the counter. Thus 16 registers of eight bits each are available.

    Each of these 16 registers is operated or selected in turn when the counter is running or being manually stepped and the contents of each can be controlled by the input switches to store either a binary 0 or 1. The actual output of each bit is inverted and is reoriented by the output gates, IC9/ 1C10.

    Besides the address lines, input lines and output lines, there is a "chip select" and a "write" line.
    

    ##  cenerator

    BY BRICE WARD
    

    Fig. 1. Block diagram of rhythm generator

    The chip select line is permanently grounded in this application because it simply makes the output available. The "write" line is used to write new information into storage when programming the rhythm generator. Setting the input switches as desired (S5 through 12) and pressing the "Write" switch will cause the new information to be stored,
    otherwise, the input does not effect the condition of the storage locations.

    The output of the instrument gates is controlled by the output of the storage section and the divide by four counter. The result is a positive pulse of voltage initiated when the counter changes state if, in the new count position, a "l" has been stored for that instrument.

    ## : Rhythms can be programmed by the user

    ## \% Eight percussion generators

    \%: Can produce any 16 pulse rhythms up to 64,000

    ## \% Four time signature switches for $4 / 4,3 / 4,6 / 8$ or $9 / 8$ measures

    ## \%: Pulse order can be observed using a Minitron Indicator

    ## PERCUSSION GENERATORS

    The percussion generators make use of various circuits to generate the sounds of drums, bongos and so forth.

    The bass drum generator, for example, contains an input amplifier which converts the 5 volt input pulse to 18 volts, shapes it and applies it to a twin-T ringing oscillator. The circuitry for the bongos and blocks is similar.

    The snare drum, cymbal and brushes make use of a noise diode the output of which is shaped and filtered, mixed with some basic frequency and applied to the mixer. The first four instruments are applied to a single f.e.t. stage. The final four are applied to individual fe.t. stages and the output of these five f.e.t.s is applied to a source follower as an output stage to the amplifier.

    The output of the 18 volt power supply is regulated to reduce hum and noise while the 4.7 V supply is held reasonably level with a Zener diode The volume control, VR16, is located on the back panel of the complete instrument.
    

    ```
    Resistors
    R1 1k\Omega
    R2 22k\Omega
    R3 15k \Omega
    R4 1k\Omega
    All 10% \frac{1}{2}```

[^5]:    
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[^6]:    Published approximately on the 15 th of each month by IPC Magazines Lid.. Fleetway House. Farringdon Street. London. E.C. 4 . Printed in England by Chapel River Press. Andover, Hants. Sole Agents for Australia and New Zealand-Gordon \& Gotch (A/sia) Lid.: South Africa-Central News Agency Lid. Publishers' Subscription Rate (including postage): Inland £3.25. Overseas £3.50.
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