PRACTICAL

ELECTRONICS

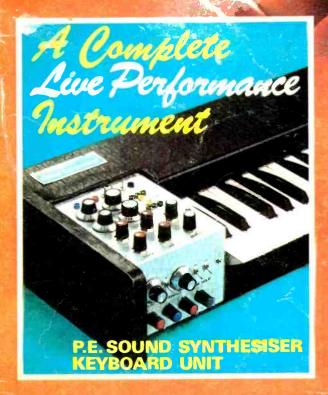
NOVEMBER 1973

20p

109

INSIDE !

OF ANY OF THE PROJECTS
FEATURED IN OUR



PAGE SUPPLEMENT

- LIGHT-OPERATED DEVICE
- WAA WAA
- DIODE THERMOMETER
- TOUCH SWITCH

STAGE LIGHTING DIMMER

is this the price you pay?

Probably if you're still using an ordinary soldering iron Ordinary soldering irons can cause damage to transistors and integrated circuits - damage which wastes time and costs money. Now, with the unique ANTEX X25 and CCN low leakage soldering irons no harm can come to the most delicate equipment, even when soldered 'Live'

(You could be making quite a saving). All prices include V.A.T. at 10%

MODEL X25 220-240 Volts or 100-120 Volts. The leakage current of the NEW X25 is only a few microamps and cannot harm the most delicate. equipment even when soldered Tested at 1500v. A.C. This 25 watt iron with it's truly remarkable heat-capacity will easily "out-solder" any conventionally-made 40 and 60 watt soldering irons, due to its unique construction.advantages. Fitted long-life ironcoated bit 1/8" other bits available 3/32" and

3/16". Totally enclosed element ceramic and steel shaft. Bits do not bits available 'freeze'' and can 1/8", 3/16" easily be removed 1/4" and 3/64" PRICE: £1.93 (rec. PRICE: £2-15 retail) P&P8p. (rec. retail). Suitable for P & P 5p.

MODEL CCN

220 voits or 240 volts. The 15 watt miniature model CCN also has nealibible leakage. Test voltage 4000v. A.C Totally enclosed element in

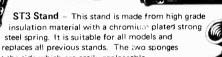
Fitted long life ironcoated bit 3/32", 4 other

ceramic shaft

MODEL G - 18 watt miniature iron, fitted with long life iron-coated bit 3/32". Voltages 240, 220 or 110. PRICE: £2-15 (rec. retail), P & P 5p.

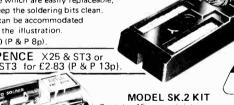


MODEL C - Miniature 15 watt soldering iron fitted 3/32" ironcoated bit. Many other bits available from 3/64" to 3/16". Voltages 240, 220, 110, 50 or 24. PRICE: £1.93 (rec. retail). P & P 5p.



at the side which are easily replaceable, serve to keep the soldering bits clean. Spare bits can be accommodated as shown on the illustration. PRICE: £1.00 (P & P 8p).

SAVE 10 PENCE X25 & ST3 or C240 & ST3 for £2.83 (P & P 13p)





production work and as a general

purpose iron.

MODEL MLX KIT

Battery-operated 12v. 25 watt iron fitted with 15' lead and 2 heavy clips for connection to car battery. Packed in strong plastic wallet with booklet 'How to Solder'

PRICE: £2.27 (rec. retail), P & P 12p

MODEL SK,1 KIT

Contains 15 watt solder, stand miniature iron fitted with 3/16' bit, 2 spare bits 5/32" and 3/32", heat sink, solder, stand and "How to Solder" booklet. PRICE: £3-25 (rec_retail). P&P12p* (rec. retail) P&P8p.

Contains 15 watt miniature iron fitted with 3/16' bit, 2 spare bits 5/32' and 3/32" heat sink,

> and "How to Solder booklet. PRICE: £2.97



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ELECTRONICS

VOLUME 9 No. 11 NOVEMBER 1973

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SPECIAL SUPPLEMENT—AUTUMN QUICKIES

DIODE THERMOMETER—WAA WAA—OPTOELECTRONICS—TOUCH SWITCH

FREE INSIDE THIS ISSUE: VEROBOARD PRINTED WIRING BOARD

Our December issue will be published on Friday, November 9, 1973

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the only way you can get so much quality for so little money!

The unique VISCOUNT III amplifier, plus the Garrard SP25 Mk III deck, plus the magnificent Duo Type III matched speakers (or Duo Type III for a small room) give you an audio installation that will prove unbeatable for listening pleasure! On the brushed aluminium front panel of the amplifier you'll find all the facilities you need—volume, bass, treble and bance controls, plus switches for mono/stereo, on/off function and bass and reble filters. Plus headphone socket on

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SPEAKERS: Duo Type II Size approx. 17in \times 10 \ddagger in \times 6 \ddagger in. Drive unit 13in \times 8in with parasitic tweeter. Max. power 10 watts 8 ohms. Simulated Teak cabinet. £14·00 a pair. £2 20 p. 8 p. Duo Type III Size approx. 23 \ddagger in \times 11 \ddagger in \times 9 \ddagger in with HF speaker. Max. power 20 watts. 8 ohms. Freq. range 20Hz to 20kHz. Teak veneer cabinet. £32·00 a pair. \pm £3·30 p. 8 p.

PRICES: SYSTEM 1

Viscount III R 102 amplifier 2 Duo Type II speakers Garrard SP25 Mk, III with £24 · 20 + £1 p & p £14 · 00 + £2 · 20 p & p MAG. cartridge plinth & cover £18-00 + £1-75 p & p

Available complete for only £49.00 + £3.50 p. & p.

 PRICES: SYSTEM 2
 2
 24 · 20 · 21 p.8 p.

 Viscount R 102 amplifier
 £24 · 20 · 21 p.8 p.

 2 Duo Type III speakers
 £32 · 00 · - £3 · 30 p.8 p.

 Garrard SP25 Mk III with MAG: cartridge plinth & cover
 £18 · 00 · + £1 · 75 p.8 p.

Available complete for £65.00 + £4 p. & p.

THE TOURIST PUSH-BUTTON CAR RADIO KIT £6-60 The Tourist PB is suitable for 12 volt working on both negative and positive earth vehicles it covers the full medium and long wave bands. It is permeability tuned and sturdily constructed. Output is a full 2.5 watts into an 8 ohms speaker. But the Tourist PB will operate into any loudspeaker from

8 to 15 ohms. Apart from the output stage, which is an integrated circuit, the only other electronic components that need soldering are some capacitors, resistors, etc. The kit includes a pre-built RF tuner unit, and fully modulised IF stages which are pre-aligned before despatch. As well as electronic components this kit also contains 2 diamond-spun aluminium knobs, elegant matching front panel, dial, washers, screws and wire.

dial, washers, screws and wire.

The Tourist PB can be mounted in any standard size dash panel and it has an illuminated tuning scale. Chassis size is: 7in wide, 2in high and 4\frac{1}{2}\text{in deep.}

* Circuit diagram and comprehensive instructions 55p free with parts.

* Fully retractable and lockable car aerial £1:37 post paid.

CAR RADIO KIT £6-60 p. & p. 55p Speaker with baffle and fixing strips £1-65. 23p p. & p., post free if bought with the kit. Send stamped addressed envelope for leaflet.

If you can solder on printed circuit board, you can build this push-button car -just follow the step-by-step instructions



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deck, less heads. Caters up to 5½ins. spools. 240V AC mains. Unused but store soiled hence no £4.00

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FIVE

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INPUTS 1. Crystal Mix or Guitar 9mV. 2. Moving coil Mic. or Guitar 8mV. Inputs 3, 4 & 5 are suitable for a wide range of medium output equipment (Gram. Tuner, Monitor, Organ, etc.). All 250mV sensitivity. Output 20 watts into 8 ohms (suitable for 15 ohms). Size approx. 12; × 6 × 3; ins. 113.50 p. & p. 60p.

UNISOUND MODULES ONLY £7.64 + 55p p. & p.

For the man who wants to design his own For the man who wants to design his own stereo—here's your chance to start, with Unisound—pre-amp, power amplifier and control panel. No soldering—just simply screw together. 4 watts per channel into 8 ohms. Inputs: 120mV (for ceramic cartridge). The heart of Unisound is high efficiency I.C. monolithic power chips which ensure very low distortion over the audio spectrum.

the audio spectrum.

IN-CAR ENTERTAINMENT

With this elegant stereo 8 track add on unit, audio enthusiasts now have the opportunity to extend their systems to include the playing of 8 track cartridges. Simply select your channel, by push button, four digital lamps indicate channel selected. Mains operated.

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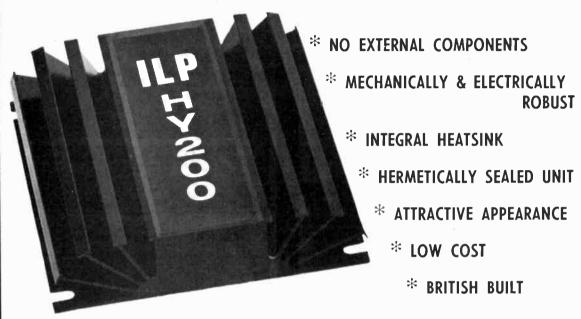
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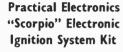
Please note we reserve the right to substitute at our discretion updated versions of advertised designs where applicable.

Mk III Sound to Light Unit Chassis Version



The audio drive voltage is derived directly from the amplifier output or across the speaker load. The unit converts the audio-frequency signals into a three-coloured light display when used with coloured flood lamps or similar; the colour depending on the frequency of the signal and the intensity on the loudness of the audio source.

Max. power 1-SkW per channel at 240V a.c.
Complete assembly built and tested. Size: 9in × 7in × 3in, Price £27-50.





This Capacitor-Discharge Electronic Ignition system was described in the November and December issues of Practical Electronics. It is suitable for incorporating in any 12V ignition system in cars, boats, go-karts, etc.

Case size: 7-25in > 4-5in × 2in approx.

Complete assembly and wiring manual 25p.

Price £12-10.



MN3 3 Channel I.C. Mixer Kit

Three channel I.C. audio mixer kit MN3 is specially Three channel I.C. audio mixer kit MN3 is specially designed for the home constructor or educational project. Suitable for use in domestic audio, disconteque and simple p.a. applications. The MN3 mixer kit has the following features: Advanced design using 5 ICs; Slider fader level controls mounted directly to p.c. board; Full range Bass and Treble controls; Uses top grade components with fibreglass ready-drilled and tinned printed circuit board and the unit may be operated from twin 9V batteries (PP3) if required.

As optional extras, an attractive ready-punched facia panel and control knobs are available. The unit is available as a ready built mixer with facia. Size: 9.5in × 4.8in × 1.5in (with facia).

Construction manual available separately at 25n. Kit Price £11.



STL/I Single Channel Sound to Light Unit

Single channel "Sound to Light" unit with 60mm slider fader controls for audio trigger level and background load dimmer.
This unit is switchable for response to High, Mid. and Low frequency audio signals, selected by facia control. A D.J. "pulse-flash" push button is fitted for manual flashing of the lamp load together with neon load indicator, Maximum Load: IkW at 240V a.c.
Size: 8in × 3:6in × 3in. Price £14-30.

STL/3 Sound to Light Unit



The performance of this unit is similar to the Mk III chassis version above, but is provided with an attractive black anodised facia panel and designed attractive black anodised facia panel and designed to mount directly into equipment desks or cabinets. High, Mid. and Low frequency channel sensitivity controls employ 60mm slider faders for present-day attractive appearance and ease of use. Size Ilin x 8-5in x 4in. Price £42-35.

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71b BARGAIN PARCELS

TIB BARGAIN PARCELS

Hundreds of new components—capacitors, resistors, switches, crystals, pots, PC boards, etc., etc. Outstanding value, £1-85 (37p).

COMPUTER PANELS: Type E: 4 0C29, 4 ACY19, 8 other transistors, 35 diodes, etc., £1 (10p). Type H: 12 ASZ20+176 R's, C's & D's, 40p (10p). Parcel of 12 top quality boards, inc. power transistors, trimpots, IC's, etc., £2 (25p); 100 for £15 (£1); 1,000 €75 (£3); 31b of lower quality, £1 (25p); 71bs, £2 (40p); 56lbs, £12 (c.pd.). Pack of boards containing at least 500 components including 50 transistors, 60p (15p).

NEW COMPONENTS: 74IC TO99 or 8 pin DIL, 32p; BC107-9, 8p or 14 for £1; 2N3055, 35p; 10+, 32p; 50+, 29p. OC140 25p; OC170 15p; BFY18 12p; BCY72 8p.

PANEL METERS: Clearing 200 meters from 10p each to callers only. Oscilloscopes available, also lots for dud units for spares. TF144G sig. gen. 85kHz-25MHz, from £12-£22. Oscilloscopes available: CD711S2 from £35. CD1212, etc. BFO No. 8 audio oscillator 50Hz-20kHz, £10. 10× crystals, 25 asstd., 75p (25p); 4p8wswitches, 40p (10p); 2-8pF beehive trimmers, 5p; 3×2‡in 15A thermocouple meters, £1 (10p); delay line unit, operating around IMHz with 7 transistor amplifier. Short delay, 40p; long delay, 40p (10p each). IMQ pots, 5p. 500 asstd. resistors £1 (15p); 20,000 £25 (£1); 100,000 £80 (£2); 300 asstd. capacitors £1 (25p).

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GREENWELD ELECTRONICS (PE5)

All mail to 24 Goodhart Way, West Wickham, Kent, BR4 0ES. Shop at 21 Deptford Broadway, SE8 (next to old cinema). Tel. 01-692 2009. Callers most welcome. Also 38 Lower Addiscombe Road, Croydon.



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- *10 Transistors *3½" Loudspeaker

NEW ROAMER

WITH V.H.F. INCLUDING **AIRCRAFT**

Transistors, Tunable wave

bands as Roamer Ten. Built in ferrite rod aerial for MW/LW

in territe rod aerial for MW/LW. Retractable chrome plated telescopic aerial for VHF and SW. Push Pull output using 600 mW transistors. 9 Transistors and 3 diodes, tuning condenser with VHF section, separate coil for aircraft, moving coil oudspeaker, volume ON/OFF and wavechange. Attractive all white case with red grille and carrying strap. Size 94 in × 7 in × 2 in approx. Parts price list and plans 30p (FREE with parts).

BUILDING

£6.95

P.P. & INS. 44p (OVERSEAS P. & P. £1·85)

POCKET

3 Tunable wavebands, MW/LW and Trawler Band. 7 stages, 5 stages, transistors and

transistors and doddes, supersensitive ferrite rod aerial, moving coil loudspeaker, attractive Black and Gold Case. Size 5½in × 1½in × 3½in approx. Plans and parts price list 15p (FREE with parts).



Total Building Costs £2.50 P.P. & Ins.

(+10% VAT 25p) (Overseas P. & P. £1·25)

Total Building Costs £2.28

(+10% VAT 22p) P.P. & Ins. 24p (Overseas P. P. £1.25)

TRANSONA FIVE

Wavebands, transistors and speaker as Pocket Five. Larger Case with Red Speaker Grille and Tuning Dial. Plans and parts price list 15p (FREE with marts) list 1 parts).

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EV5 5 Transistors and 2 diodes. MW/LW.
Powered by 4½ voit Battery. Ferrite rod aerial, tuning condenser, volume control, and loudspeaker. Attractive case with red speaker grille. Size 9in x 5½in x 2½in approx. Parts price list and plans 15p (FREE with parts).

TOTAL P.P. & INC 200-P.P. & INS. 30p (OVERSEAS P. & P. £1·25)

BUILDING

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ROAMER EIGHT Mk. I

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7 TUNABLE WAVEBANDS:

7 TUNABLE WAVEBANDS:

MW1, MW2, LW, SW1,

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10 TAL BUILDING COSTS.

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FOR EASIER TURING OF LUXEMBOURG, ETC. Sensitive territe rou action and Secopic aerial for short waves. 3in speaker. 8 stages—6 transistors and 2 diodes, etc. Attractive black case with red grifle, dial and black knobs with polished metal lineeric. Size 9in x 5 in x 2 in approx. Plans and parts price list 25p (FREE with parts).

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SWI, SW2, SW3,
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STATIONS. ALSO AIRCRAFT BAND
STATIONS. ALSO AIRCRAFT BAND
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tetractable, chrome piated 7 section telescopic
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P.P. & INS. 52p

TOTAL

BUILDING COSTS

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× 2\in	74p	55p	41p	10o		
× 3≟in	99p	77p	57 Lp	10p		
× 5in	_	_	82 p	10p		
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BAX13 BAX16 BAY38 BB105 2N4289 2N4441 2N4442 2N4443 2N4444 2N5129 2N5172 2N5577 2N6076 3N140 3N152 3N153 40360 AAZ13 BA100 BA102 BA110 BA111 BA112 BA114 BA115 BA128 BA128 BA128 BA148 BA148 BA148 BA154 BA154 BA155 BA166 BA182 BA182 BA182 BA182 BA221 BA222 BA221 BA221 11p 10p 27±p 45p 55p 9p 9p 11p 23p 75p 22p 22p 22p 7p 7p 7ip 11p 9p 10p 41p 42p 99p 98p 94p 11p 10p 8p 8p 9p 11p 12p OA10 OA47 OA79 OA81 OA85 OA90 OA91 OA200 OA202 OA210 OA211 36 i p 47 p 14 p 11 p 10 1 p 89 p 10 1 p 50 p 50 p 50 p 50 p 51 · 39 £2 · 17 55 p 10 p 10 p 28 p 99p 16ip 19p 23p 25p 13ip

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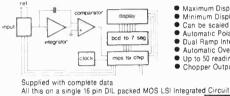
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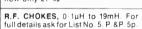
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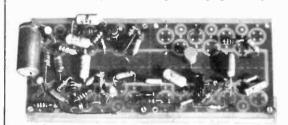
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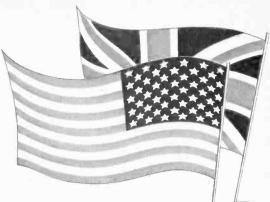
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8N7423N	87	34	32	8N7494N	85	80	75	8N74177N	1.44	1.44	1.26
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8N7428N	48	48	87	8N74100N	2.16	1.89	1.89	8N74182N	1.44	1.44	1.26
8N7430N	20	18	16	8N74104N	60	53	45	8N74184N	2.16	2.16	1.89
8N7432N	87	87	32	8N74105N	60	53	45	SN74185AN		2-16	1-89
8N7433N	48	48	88	8N74107N	51	51	45	8N74188N	6-48	6.48	5.67
8N7433AN	57	57	50	SN74110N	57	57	50	SN74190N	2.80	2-80	2-01
8N7437N	48	48	87	8N74111N	86	86	75	8N74191N	2.30	2-80	2-01
SN 7438 N	48	48	87	8N74116N	2-16	2.16	1.89	8N74192N	2-80	2 80	2.01
SN7438AN	57	57	50	8 N74118N	1.00	90	88	SN74193N	2.30	2-80	2.01
8N7440N	20	18	16	8N74119N	1.92	1-92	1-68	8N74194N	1.72	1.72	1.51
8N7441AN	85	79	78	8N74120N	1.05	1.05	92	8N74195N	1-44	1-44	1.26
8N7442N	85	79	78	8N74121N	57	57	50	8N74196N	1-58	1.58	1.38
8N7443N	1.50	1.27	1-18	8N74122N	80	80	70	8N74197N	1.58	1.58	1.88
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SN74 (HIGI	H PO	WER), SN	74 (LOW F	OWE	R) se	ries i	in stock. Se	end fo	r Lis	t 36
free on req	uest.	Low	Profi	le Sockets,	14 pin	, I5p	, 16 p	in, 17p, 8 p	in, 14	p.	
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2N697 2N698	0·15 0·30	BC109	U-12	DD000	0.10	OC24 OC25	0-40	7403	0.20
2N706	0.10	BC113	0.16	DD008	0.88	OC26	0.85	7404 7405	0.20
2N706A	0.12	BC115 BC116	0.20 0.20	GD3 GD4	0.88	OC28 OC29	0-65 0-65	7406	0-40
2N708 2N709	0-10	BC116A	0.23	GD5	0-10 0-88	OC30	0.40	7407	0-40
2N1091	0.55	BC118 BC121	0-15	GD8	0-25	OC35	0.55	7408 7409	0-25 0-88
2N1131	0.25	BC121 BC122	0-20 0-20	GD12 GET102	0·10 0·50	OC36 OC41	0-65 0-85	7410	0.20
2N1132 2N1302	0·25 0·18	BC125	0.68	GET103	0-40	OC42	0.40	7411	0.28
2N 1303	0.18	BC126 BC140	0.65	GET113	0.85	OC43	0.70	7412 7413	0.28 0.30
2N1304	0.22	BC140 BC147	0.55 0.12	GET114 GET115	0-80 0-75	OC44 OC44M	0·18 0·17	7416	0.80
2N1305 2N1306	0.22 0.28	BC148	0.10	GET116	0.75	OC45	0.18	7417	0.30
2N1307	0.28	BC148 BC149	0.12	GET120	0.50	OC45M OC46	0.18	7420 7422	0.20
2N1308	0.28	BC157 BC158	0·14 0·12	GET872	0.80	OC48 OC57	0-27 0-60	7423	0.40
2N2147 2N148	0·75 0·60	BC160	0-63	GET875 GET880	0.40 0.55	OC58	0.60	7425	0.87
2N2160	0.61	BC169	0-14	GET881	0.25	OC59	0.60	7427 7428	0.87
2N2218	0.28	BCY31	0.45	GET882	0-85	OC66	0.50	7430	0.43
2N2219 2N2369A	0.25	BCY32 BCY33	0.88	GET885 GEX44	0.35	0C70 0C71	0·18 0·15	7432	0.87
2N2444	1.99	BCY34	0.45	GEX45/1	0.45	OC72	0.25	7433	0.48
2N2613	0.28	BCY38	0.55 1.00	GEX941 GJ3M	0.45	I OC73	0.50 0.80	7437 7438	0-48 0-48
2N2646 2N2904	0-50 0-20	BCY39 BCY40	0.80	GJ3M GJ4M	0.50 0.38	0C74 0C75	0.80	7440	0.20
2N2904A		RCV49	0.80	GJ5M	0.25	OC76	0.30	7441AN	0.85
2N2906	0.20	BCY70 BCY71 BCZ10	0-15 0-20	GJ7M	0.50	OC77	0-55	7442 7450	0-85 0-20
2N2907 2N2924	0.23 0.28	BCZ10	0.60	HG1005 H8100A	0.50	OC78 OC79	0-25 0-80	7451	0.20
2N2925	0.15	BCZ11	0.65	MAT100	0.20	OC81	0.28	7453	0.20
2N2926	0.10	BD121	1.00 1.00	MAT101	0.25	OC81D	0.28	7454 7460	0.20
2N3054	0.45	BD123 BD124	0.80	MAT120 MAT121	0-20 0-25	OC81M OC81DM	0.20 0.18	7470	0.33
2N3055 2N3702	0.11	BDY11	1-45	MJE520	0.65	OC81Z	0.45	7472	0.88
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2N3706	0.11	BF117 BF167	0.50 0.28	MJE3055 NKT128	0-75 0-45	OC82D OC83	0·25 0·25	7475	0.59
2N3707 2N3709	0.10	BF173	0-25	NKT129	0-80	OC84	0.25	7476	0.45
2N3710	0.11	BF181 BF184	0.88	NKT211 NKT213	0.25	OC114	0.88	7480 7482	0.80 0.87
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28301 28304	0.59 1.15	BF196 BF197	0.15	NKT218 NKT219	1.13 0.83	OC141 OC169	0.80 0.20	741AN	1.10
28501	0-87	BF861	0.25	NKT219 NKT222	0.80	OC170	0-25	7492 7493	0-75 0-75
28703	1.00	BF898 BFX12	0.25 0.20	NKT224	0.25	OC171 OC200 OC201	0.30 0.55	7493	0.75
AA129 AAZ12	0-20 0-75	BFX13	0.25	NKT251 NKT271	0.24 0.20	OC201	0.80	7495	0.85
AAZ13	n-10	BFX29	0.28	NKT272	0.20	OC202	0-90	7496 7497	1.00 4.82
AC107	0.85	BFX30 BFX35	0.28	NKT273 NKT274	0.20	OC203 OC204	0-55 0-65	74100	2.16
AC126 AC127	0-25	BFX63	0.50	NKT274	0-20 0-25	OC205	1.00	74107	0.51
AC128	0.20	BFX84	0.25	NKT277	0.20	OC206 OC207	1.10	74110 74111	0-57 0-86
AC187 AC188	0.20	BFX85 BFX86	0.28 0.25	NKT278 NKT301	0.25	OC207 OC460	1.00 0.20	74118	1.00
ACV17	0.85	BFX87	0-25	NKT304	0.85 0.75	OC470	0.80	74119	1.92
ACY18 ACY19	0.27	BFX88	0.22	NKT403 NKT404	0.70	OCP71	1.00	74121 74122	0.57 0.80
ACY19 ACY20	0-27 0-22	BFY10	1.00 1.25	NKT404 NKT678	0.60	ORP12 ORP60	0.55 0.45	74123	1.44
ACY21	0.22	BFY11 BFY17	0.25	NKT713	0.80	ORP61	0.48	74141	1.00
ACY21 ACY22	0.16	BFY18	0-45	NKT713 NKT773	0.25	819T	0.80	74145 74150	1·44 2·80
ACY27 ACY28	0-25	BFY19 BFY24	0-55 0-45	NKT777 078B	0-88 0-88	8AC40 8FT308	0.25	74151	1.15
ACY39	0-65	BFY44	1.00	OA6	0.12	ST722	0.88	74154	2.80
ACY39 ACY40	0.22	BFY50	0.20	OA47	0.08	ST7231	0.68	74155 74156	1.15
ACY41 ACY44	0.22	BFY51 BFY52	0·20 0·20	OA70 OA71	0·10 0·10	8X68 8X631	0.20 0.45	74157	1.09
AD140	0.50	BFY53	0.17	OA71	0.15	8X635	0.55	74170	2.88
AD149	0.20	BFY64	0.45	OA74	0-15	SX 640	0.75	74174	1.80
AD161	0-89	BFY90 BSX27	0.75 0.50	OA79	0.10	8X641 8X642	0.75 0.60	74175	1.29
AD162 AF106	0.89	BSX60	0.98	OA81 OA85	0-10 0-15	SX644	0.85	74176	1.44
AF114	0.25	BSX76	0-18	OA86	0.15	SX 645	0.85	74190	2-30 2-30
AF115 AF116	0.25	BSY26 BSY27	0·18 0·18	OA90 OA91	0.07	V15/30P V30/201P	0-75 0-75	74191 74192	2.80
AF116 AF117	0-20	BSY51	0.50	OA91 OA95	0.07	V 60/201	0.50	74192	2.30
AF118	0.50	B8Y95A B8Y95	0.12	OA200	0.08	V60/201P	0.75	74194	1.72
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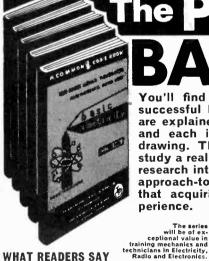


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2.2	_	_	_	_	10p	_	7p	8p
4.7	_	_	_	10p	_	7p	8p	7p
10	_	_	_	_	7p	8p	7p	8p
22		-	7p	7p	_	7 _D	7p	9p
47	7p	_	8p	8p	8p	7 p	9p	120
100	8p	7p	7p	7p	7p	9p	Hp	19p
220	7p	8p	8p	8p	9p	100	17p	27p
470	8p	9p	9p	100	12p	17p	24p	43p
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C		3/0		E12	9
C	1/8W	5%	4-7 Ω –470Κ Ω	E24	1
C	1/4W	5%	4·7 Ω = 10M Ω	EI2	i
C	1/2W	5% 5%	4-7 Ω-10M Ω	E24	i
C C C	IW	5%	4-7 Ω - 10M Ω	ĒĪŽ	,
MO	1/2W	2%	10 Ω – IM Ω	E24	- 2
ww	IW	$10\% \pm 1/20 \Omega$	0.22 Ω -3.9 Ω	E12	7
ww	3W	5%	Ι Ω-ΙΟΚ Ω	E12	,
ww	7W	5%	Ι Ω-ΙΟΚ Ω	E12	ý
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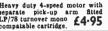
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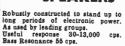
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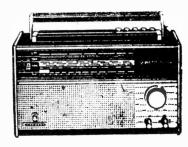
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100

178

183

0-10

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AL10/AL20/AL30 **AUDIO AMPLIFIER MODULES**



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The versatility of their design makes then ideal for use in record players, tape recorders, stereo amplifiers and cassette and cartridge tape players in the car and at home.

Parameter	Conditions	Performance
HARMONIC DISTORTION	Po = 3 WATTS f = 1KHz	0.25%
LOAD IMPEDANCE		8−16Ω
INPUT IMPEDANCE	f = 1KHz	100 kΩ
FREQUENCY RESPONSE -3dB	Po = 2 WATTS	50 Hz-25KHz
SENSITIVITY for RATED O/P	Vs=25V, R1=80 f=1KHz	75mV. RM8
DIMENSIONS		3" × 21" = 1"

The above table relates to the AL10, AL20 and AL30 modules. The following table outlines the differences in their working conditions.

Parameter	AL10	AL20	AL80
Maximum Supply Voltage	25	30	30
Power out for 2% T.H.D. (RL = 8Ω f = 1KHz)	3 watts RMS Min.	5 watts RMS Min.	10 watts RM8 Min.

AUDIO AMPLIFIER MODULES

3 watts 5 watts 10 watts

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20 Hz-50 K Hz (-3dD)
Bass control—
± 12dB at 60Hz
Treble control—

Treble control—

± 14dB at 14KHz
*Input 1. Impedance
1 Meg. ohn
Sensitivity 300mV
†Input 2. Impedance
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Sensitivity 4mV

EA1000 AUDIO AMP MODULE 5 WATTS R.M.S.

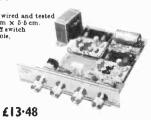


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50W pk 25w (RMS)

0.1% DISTORTION! HI-FI AUDIO AMPLIFIER

THE ALSO

- Frequency Response 15Hz to 100,000-1dB.
- ★ Load-3, 4, 8 or 16 ohms,
- ★ Distortion-better than .1% at 1KHz
- ★ Signal to noise ratio 80dB.

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- ★ Supply voltage 10-35 Volts.
- ★ Overall size 63mm 105mm × 13mm.

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Filters: Rumble (High Pass)
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Signal/Noise Ratio

Input overload Supply Dimensions

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TAKE YOUR PICK

The tendency towards larger and more ambitious projects is a perfectly natural development in the wake of technological advancement. Not everyone wants to become involved in building the more complex project, that is true. But this expansive side of home construction should be welcomed by all the genuine enthusiasts, since it provides an opportunity for amateurs to keep up to date with the latest developments and techniques and to make practical use of them as and when they feel so inclined.

It is worth considering, also, the beneficial effect on the retail components market arising from the demand generated by some larger projects. Such demands encourage suppliers to hold a wide variety of stocks and to seek out lines in components they might not otherwise contemplate handling. A healthy and venturesome retail trade is to the advantage of everyone who purchases components, whether in large or small quantities.

The needs and interests of constructors vary widely, but so far as generalisation can go, it seems reasonable to suppose that out of the whole, a majority of constructors confine their activities to the more simple kind of project. Newcomers undoubtedly start this way too, although many will in due course feel the compulsion to pursue their hobby to greater heights. Our recognition of the need for small simple designs is reflected in this month's special supplement. This contains details of four uncomplicated but useful and highly adaptable circuits, any of which can be assembled on the piece of printed wiring board presented free with this issue of Practical Electronics.

Returning now to larger projects, it may be recalled that the virtues of modular construction were discussed here last February when the P.E. Sound Synthesiser made its debut. With the synthesiser main unit now completed, this month's article in the series introduces the complementary keyboard unit.

It will interest many readers to learn that over the past 12 months the complete synthesiser has been handled by a number of professional musicians, and its unique creative possibilities—both as a composing machine and as a live performance instrument—have been established and enthused upon. Moreover, it has been confirmed that the keyboard unit on its own does constitute a valuable and versatile instrument for live performance. For example, it will meet the needs of many musical groups, being able to provide novel avant-garde accompaniment.

This feasibility is another bonus or "spin off" from the modular approach. It confirms the importance of considering both the *whole* and the *parts*, where many large and ambitious projects are concerned.

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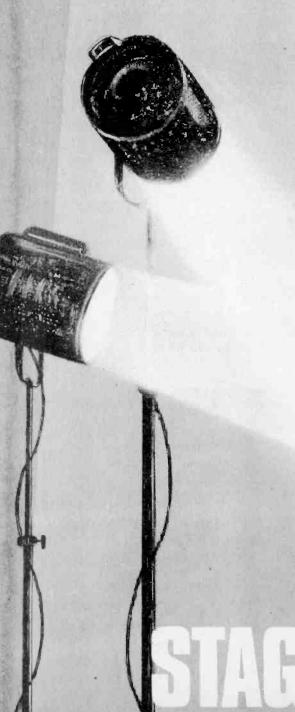
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FOUR CHANNELS OF INDEPENDENT CONTROL . . .



HIS Stage Lighting Dimmer was built to the specific requirements of an amateur revue company. It consists of four independent channels each capable of dimming 13 amps (3250 watts) of incandescent (not fluorescent) lighting. The number of channels may be increased as necessary providing the current capability of the input is uprated.

SYSTEM DESCRIPTION

Each of the four dimmable channels is controlled by a linear slider fader. Any channel can be switched to a Master Fader which is a double unit made up of two stereo slider potentiometers. The Master has control over all channels simultaneously when required.

Each channel has a mimic light next to the fader control which gives a visible representation of the relative brilliance of the lights plugged into that

A preset potentiometer in each channel allows the cutoff point of each channel to be adjusted for different wattage lamps with differing thermal capacities.

Interference suppression is included and the dimmers are mounted in an earthed box to reduce

radiation.

The output to each channel is fed via a 13 amp switched socket and input from the mains is taken via a 60 amp fused switch which can act as a "Master Blackout" control when it is necessary to cut out all the lamps in the fastest possible time.

CIRCUIT OPERATION

Each of the four channels uses an identical circuit which uses a triac to control the current through the lamps. The basis of this circuit is shown in Fig. 1.

A triac is a four layer semiconductor device with three terminals. It will only conduct when a voltage is applied to the gate (g) terminal and will then continue conducting in either direction even when the gate voltage is removed. It will cease to conduct when the current through it drops below a threshold level known as the holding current.

The circuit shown in Fig. 1 varies the average current through the lamps by controlling the point in the mains cycle at which the triac is triggered into conduction. This method of control is known as phase control.

EACH CAPABLE OF DIMMING 31 kW OF LIGHTING

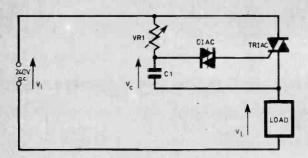


Fig. 1. The basic phase control triac circuit. The waveforms in different parts of the circuit are shown right

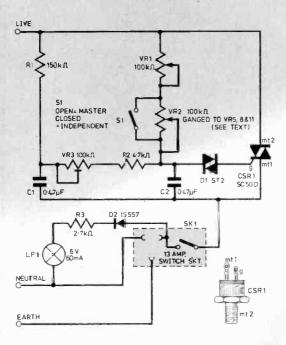
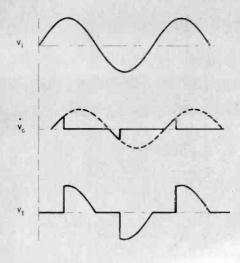


Fig. 2. The circuit of one of the control channels of the Stage Lighting Dimmer. Four identical channels are used in the unit



Referring to Fig. 1, capacitor C1 and variable resistor VR1 form a phase shift circuit which alters the phase of the signal at their junction with respect to the applied mains voltage.

The diac in series with the triac gate is a fourlayer device which presents a high impedance until the voltage across it exceeds a value known as the breakdown voltage (about 30 volts) when it presents a low impedance. When the breakdown voltage is reached the capacitor C1 discharges into the gate of the triac causing it to conduct. Typical waveforms are also shown in Fig. 1.

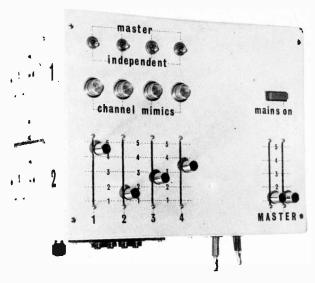
COMPLETE CIRCUIT

In the actual circuit used, shown in Fig. 2, VR1 is the Channel Fader and VR2 is one section of the Master Fader which can be over-ridden by S1 to give independent control. VR1 and C2 form the phase control network and R1 and C1 are additions to provide smooth control at low light levels.

The current supplied by R1 and C1 is determined by VR3 in series with R2. VR3 can be used to

DIWINER

By R. Liffen



Photograph showing the layout of the components on the front panel and the output sockets mounted on the side of the case. Another two sockets are similarly fixed the other side

preset the extinguishing point of the lamp when the Channel Fader is fully down. There are two modes of use for VR3: it may be used to fully extinguish a lamp which would otherwise "sing" (due to the filament vibrating in the presence of small residual spiky current; alternatively it may be set so that the filament glows dull red, which enables a big lamp to be taken to full brightness quickly.

MIMIC CIRCUIT

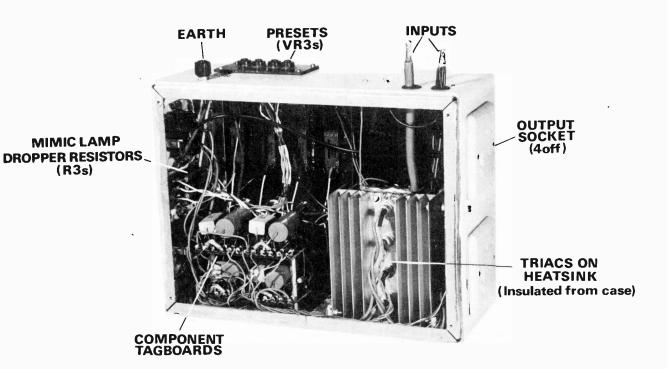
Components D2, R3 and LP1 form a "mimic" circuit to give a visual indication of the lamp brightness. D2 and R3 simply reduce the mains voltage down to six volts. The four resistors (R3 and its counterparts) dissipate nearly 10 watts each and must be placed in a well ventilated position.

Note the special connection between the switchboard 13 amp socket and D2. This is necessary because with no lighting load plugged in, the leakage through the triac will give a glow in the mimic lamp which can be disturbing to the lighting operator. With no lamp plugged in, the channel should be switched off at the socket.

The point to connect D2 will be found as a copper rivet head on the rear of the socket and should be verified with an ohmmeter.

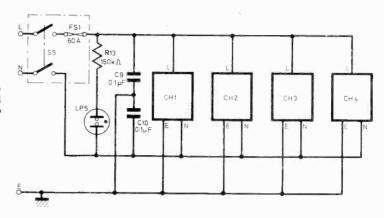
LOAD CONNECTION

The load is connected on the neutral side of the triac. This reduces the risk of electric shock from



Photograph showing the arrangement of the triac heatsink, the presets, the tagboards and the input connections within the case. The heatsink must not touch the earthed case

Fig. 3. Block diagram showing the four circuits of Fig. 2 interconnected and the extra components needed to complete the unit



defective lamps, since the lamp is at neutral rather than live when faded down. It also enables all the triacs to be mounted on a heatsink without insulating washers although this has the disadvantage that the heatsink is live and must therefore be insulated from the case.

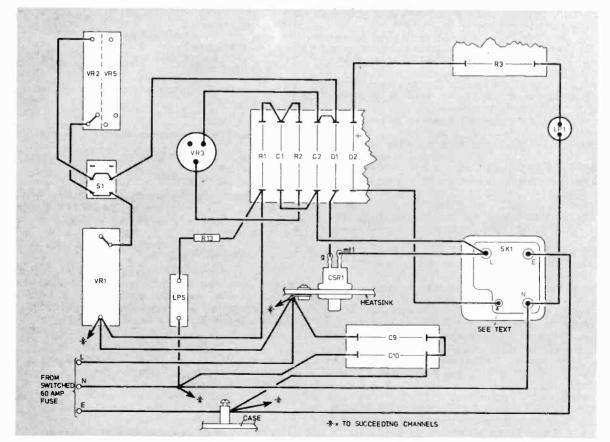
The triac, though a bi-directional device, operates best with the main terminal 2 (mt2) live with respect to mtl. For the SC50D mt2 is the threaded stud.

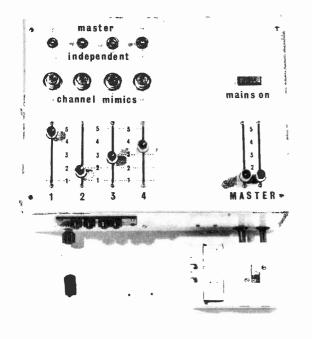
Interference suppression is obtained using two capacitors (C9 and C10) connected between the supply lines as shown in Fig. 3. Also seen on Fig. 3 is the mains indicator neon and the 60 amp fused switch.

CONSTRUCTION

The dimmer was designed to be mounted on a wall so the control box and the switched fuse were

Fig. 4. Layout of the components on the tagboard and interwiring for one of the four channels. Components should be mounted on the front panel, the case and the tagboards, and then interwiring should be done using this diagram and Fig. 3 as a guide. Wires which carry the current to the lamps should be heavy duty to reduce heating effects.





Photograph showing the completed unit and 60 amp switched fuse mounted on the backing board

fitted on a wooden panel fitted with mounting brackets (see photograph). The metal box used for the prototype is not generally available but any aluminium or steel box of sufficient size $(13\text{in} \times 10\text{in} \times 5\text{in})$ can be used.

The four triacs are mounted on a large heatsink which is connected to the live input via the switched fuse. A smear of heat-conducting grease should be applied before the triacs are bolted in position.

All the small components are mounted on two tagboards as shown in Fig. 4. The only exceptions are the four mimic lamp resistors (R3 etc.) which are mounted on a separate tagboard in a ventilated position.

The m.e.s. lampholders must be of the type which have both lamp terminals insulated from the case as neither must touch the case which is earthed.

The four preset potentiometers (VR3 etc.) are mounted on the outside of the case so that they are easily adjusted.

The output sockets are mounted two on either side of the case. The tagboards are mounted on pillars.

Plastic brackets are used to support the heatsink without it touching the case. Make sure that the lid of the case does not touch it when fitted.

Vents must be cut in the case to allow air to circulate if none are present.

Front panel layout is not critical, the photographs showing the prototype layout. Note that the two stereo sliders which form the Master Fader should be mounted as close as possible to each other so that they may be moved together. They may be physically joined with a metal bracket if this is desired.

Cables capable of carrying the 60 amp input current must be used for connection to the unit. A heavy duty earth cable should be connected to the case using a screw terminal.

COMPONENTS . . .

Resistors

R1 150k Ω R2 4·7k Ω $\frac{1}{2}$ watt carbon R3 2·7k Ω 10W wirewound

Potentiometers

VR1 $100k\Omega$ linear slider

VR2 100k Ω + 100k Ω linear stereo slider (each channel uses half of one)

Capacitors

C1, C2 0.47µF 400V (2 off)

Semiconductors

D1 Diac type ST2 (Henry's Radio)
D2 1S557, BY100 or any 400V 1A diode
CSR1 SC50D 400V 15A Triac (Henry's)

Miscellaneous

LP1 6V 60mA m.e.s. lamp and holder SK1 13A switched mains socket (MK2957) + mounting plate MK2200 21L

S1 Mains on/off switch

The components above are required for each of the four channels, i.e. four of each (except VR2) are required. The components below are required to complete the unit

Resistor

R13 150k Ω (not needed if included with neon)

Capacitors

C9, C10 0·1μF 1000V (2 off)

Miscellaneous

LP5 Mains neon indicator
S5 M.E.M. 60A switch/fuse
12 way tagboards (2 off)
4 way tagboard
Screw terminal, 6in × 6in × 2in finned

Screw terminal, $6\text{in} \times 6\text{in} \times 2\text{in}$ finned heatsink, plastic brackets, 16mm copper double insulated (p.v.c./p.v.c.)

PRACTICAL POINTS

To reduce the interference effects of the unit it should be used on a supply separate from that being used by any microphone or musical instrument amplifiers, i.e. the 60 amp cables to the input should be taken to the nearest low impedance mains supply, the local "main feeder."

Ordinary household ring main circuits will not handle the unit on full load. In the case of a house or small hall the connections should be made direct to the fuse box, having first checked that the incoming mains can supply 13kW.

Footlights present a special problem since they are often close to stage microphone cables. If possible use footlight feeder cable which has an earth screen totally enclosing the conductors.

The slots in the box for the faders was made using a Monodex metal cutter.

A heavy duty soldering iron is necessary for soldering the 60 amp cables.

Total cost of the unit was approximately £35.





BLACK HOLES

There are many enquiries about "black holes" and not surprisingly they are mostly of the form "What is a black hole?"

The term "hole" is the main difficulty, just as it was with solid state devices. Perhaps this is because usually a hole is some space which is at an angle to a surface.

Since the situation is one of concentration of matter and energy it is better to use the word concentration. The new name then is a "black concentration" and in this case it means that energy and matter is not radiating outwards but is concentrated away from the observer. It is, therefore, a black condition.

Proceeding from there, imagine that the Sun in following its life course goes through the process of normal star evolution and death. It will expand to many times its present size to become a Red Giant. In expanding it will cool down and as the fuel of it becomes consumed the gravitational forces will cause it to contract to about one per cent of its volume. This will mean that it will become very hot again. It will then acquire a new name, that of White Dwarf. Its brightness will have increased and so will its density so that a cubic inch would weigh many tons.

However, by the nature of the process there is a limit to the compression and in the case of the Sun it will probably remain as a White Dwarf, gradually cooling down.

GRAVITATIONAL COLLAPSE

In the case of stars more massive than the Sun the energy produced during the life cycle may be such as to make the stage from Red Giant to White Dwarf much more concentrated. In this case a state of gravitational collapse may occur and a "black concentration" will take place.

Not only does this effect apply to the star itself but also to the surrounding media. In this condition it is possible that other concentrations of matter nearby are sucked into the area.

The advent of gravitational collapse is likely to arise after the transition from Red Giant to White Dwarf. This could be because the White Dwarf stage is not a halting point as it were, and the energy of the shrinking from Red Giant may over-run the White Dwarf point. In this case the enormous energies will build up to a catastrophic explosion and a supernova would be the result. This will appear as an extremely bright object and parts of the original star will be flung out at fantastic speeds of the order of hundreds of miles a second. There will be left a small highly concentrated mass, the neutron star. This is the Pulsar according to the accepted current theory.

Sometimes the energy is so great that the situation is not ended at this point and the collapse goes beyond the neutron star point. It is here that the condition of the "black hole" is reached.

It is not easy to accept the implications of this manner of events. It would mean that no light or radiation of any kind comes out of this

concentration.

The gravitational field is so great that in the comparatively short distance involved (if the Sun became a "black hole" it would be only two miles in diameter) matter in moving inwards from the "surface" would stretch out from the observer. In fact, there would persist for some time an image of the original matter and this would not appear to change for a very long time. In effect there would be a reversal of all the conditions that are regarded as normal.

NEW LOOK

At this point it is time to ask whether physicists have to take a new look at the whole situation. It would seem that there are no universal laws, but rather that the laws depend upon conditions as they are found. This is not merely a matter of reality but also of profound philosophical implications. It may be that other views will be expressed and that this interpretation is in itself rejected.

There is still no direct evidence, for the existence of the "black hole" but work which is going on at the moment may help to clear up the matter. It is hoped that there will be sufficient data shortly to report on the progress of a new project aimed at providing new evidence.

COMET KOHOUTEK

The comet Kohoutek, which is likely to make a sensational appearance and will be visible to the naked eye for four months or so, will be at its peak in December when it will be visible in full daylight. It will be visible to the naked eye from about the first of November.

There was not time for the organisation of a probe to intercept this comet though in other fields the early warning of its advent is useful. Professional gatherings have been held to set up a study programme and to avoid unnecessary duplication of observations.

Studies will be made in infra-red and observations in the 50-100 micrometre wavelengths will be made from a jet plane. It may well be that a programme will be set for observations from Skylab.

It so happens that a number of dispersed observations from automatic satellites and a probe will be able to provide useful data. Satellites OSO-3 and OSO-7 will be able to make photographic observations and Mariner 10 will be able to monitor the inter-planetary medium as it proceeds on its journey to Venus and Mercury.

All these study programmes will help to practice different techniques which can be used for the special observations of Halley's comet which is due in 1986.

LIGHTS IN THE SKY

Three years ago a search for fluorescent pulses, induced by X-ray photons, was made. During this observation the fast atmospheric pulsations of light were discovered. Since this discovery there have been observations which threw more light on these effects.

At Ankara H. Ögelman has made observations with baselines of 175 and 3,300 kilometres. The results of this work are being correlated, but first results seem to indicate that these FAP's, as they are called, appear from discrete directions and are more frequent than was supposed.

The energies are being measured and later results will appear in this column.

MARS VEHICLES

Considerable re-design may be necessary for the manned vehicles to Mars because of the high radiation levels that might be normally encountered.

This conclusion arises from the examination of the materials of the Apollo spacecraft after returning from the Moon. The extent of the radiation has supplied the solution and it is thought a re-arrangement of the present design will be possible to overcome this hazard.

PART 3

BY R.A.COLE

THE DISCERNING CONSTRUCTORS CHOICE:

- * A Total Quadraphonic System
- * An advanced design AM/FM radio
- ★ Complete loudspeaker enclosure design
- ★ Flexible design allows individual modules to be incorporated in existing equipment
- * Switching facilities and modules for SQ, QS(RM) and CD4 systems

HE previous articles described the construction of the SQ decoder, the power amplifiers, pre-amplifiers and volume/tone/balance controls, their circuits and operation.

The present article is devoted to the construction of the power supply and the construction and assembly of the main chassis member.

POWER SUPPLY

The power requirements for the Rondo are dual rails providing positive and negative supplies with a centre earth for the preamplifiers, tone controls and power amplifiers, and a positive rail and negative earth supply for the tuners, stereo decoder and quadraphonic decoder(s).

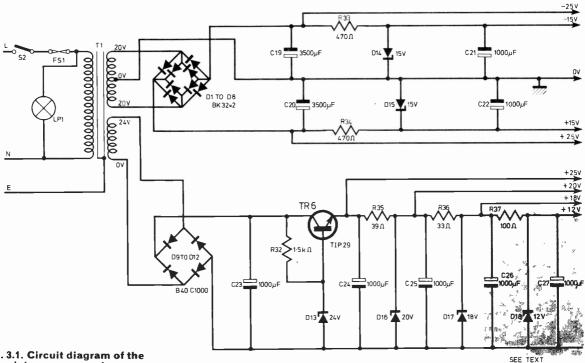


Fig. 3.1. Circuit diagram of the complete power supply

> Practical Electronics November 1973

POWER SUPPLY **BOARD**

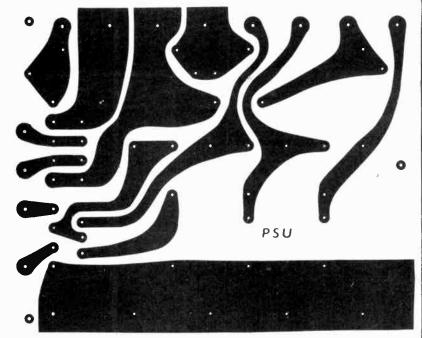


Fig. 3.2. Circuit board master for the Rondo power supply

COMPONENTS . . .

POWER SUPPLY

Resistors

R32 1.5kΩ

R33 470Ω

470Ω R34

R35 39Ω

R36 33Ω All 1W 10%.

Capacitors

C19 40 V W $3,500 \mu F$ C20 $3,500 \mu F$ 40 V W

C21 $1,000 \mu F$ 16VW

1,000μF 16 V W C22

C23 1,000 µF 35 V W

C24 1,000 µF 25 V W

C25 1,000μF 25 V W C26 1,000µF 25 V W

All elect.

Semiconductors

D1 to 8 **BK32**

(2 in parallel) Semicron.

D9 to 12 B40 C1000

Semicron.

D13 24V Zener. D14 & 15

15V 10%. Zener. 20V Zener. D16

18V Zener. D17

All Zeners are 400mW

TIP29 TR6

Transformer

Primary 240V a.c. Secondaries

1, 20-0-20V a.c., 4A.

2, 24V a.c. 1/2A.

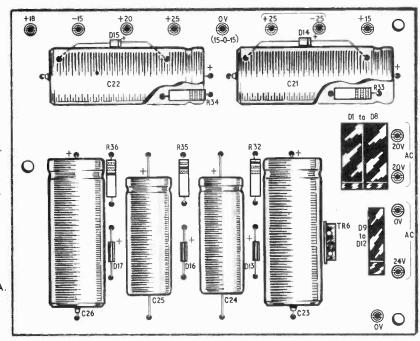


Fig. 3.3. Component layout for the power supply board viewed from the component side



MAIN CHASSIS DETAILS This Screen mounted by Transformer into 3.5. **Transformer** Fig. screen metalwork details 3/8dia grommet in slot Fig. 3.6. Power supply 89 screen metalwork details 113 This area occupied by Power Amplifier Heat Sinks. Amplifier boards and Heat Sinks are assembled and fitted in Chassis prior to installation of Power Supply Screening 4 rubber mounting feet Screen secured with 4-1/g dia PK self-tapping screws, 251 (Outside dimn) (Outside dimn) 50 Outside dimn) '9 Holes 'A' 6BA clear 9 Holes A OBA Cited (see Power Supply Screen) 29 Holes A 31-32 dia (see Power Supply Screen) 3 Holes C 3/8 dia 1 Hole D 42 dia 1 All dimensions in m.m. Material: 205WG Mild Steel f Group E 3/8 dia X 2 - 6 BA clear 6 Groups F 5/8 dia X 2 - 6 BA clear NOTE: Main dimensions only are given. Detailed dispersion of holes to be 1 Group G 3/4 dia X 2-6 BA clear made to suit individual items and components to be mounted on Chassis. Fig. 3.4. Main supporting chassis metalwork shown in general detail in the fabricated form and with external critical dimensions for simplicity. There should be no reinforcement where the

sides meet, as the sides must be able to be pressed inwards when the wooden sleeve is fitted

POWER AMPLIFIERS SUPPLY

The dual rail supply is, as shown in Fig. 3.1, derived from a 20-0-20V transformer T1 winding with a 4A capacity rectified by two bridge rectifiers D1-D8 in parallel. The expedient of using two smaller (at lower cost) unmatched, bridge rectifiers in parallel is open to criticism because of "current hogging". That is, that the rectifier with the lower inpedance will have a higher current flow. The rectifiers have been chosen so that, under all operating conditions, there are no detrimental effects due to this phenomenon.

The transformer centre tap is taken direct to the junction of two $3,500\mu\text{F}$ main smoothing capacitors C19, C20, joined negative to positive, and the rectified d.c. supply to the free positive and negative terminals of these capacitors respectively. The d.c. voltage, after rectification, is around +25 to -25V. (The centre zero is at earth potential.) The supplies to the main power amplifier output stages are taken direct from the main smoothing capacitors.

OTHER BOARD SUPPLIES

All the 748's are fed from $\pm 15 V$ d.c. rails obtained by dropping through resistors R33, R34 from the $\pm 25 V$ supplies smoothed by two $1,000 \mu F$ capacitors C21, C22. As these rails do not have to be at precisely 15V but the i.c.s have to be protected against exceeding a maximum rating of 18V, they are "clamped" to 15V by two Zener diodes (D14 and D15).

The positive rail and negative earth supply is obtained from a separate winding on the transformer T1. This winding is 24V a.c. at 250mA and is rectified by a 1A bridge rectifier, D9 to D12, smoothed and then regulated by a series regulator circuit TR6, R32 and D13 to 24V d.c. and smoothed again.

From this point is taken the supply to the varicap diodes of the f.m. tuner.

A further ladder of droppers, Zeners and smoothing capacitors produces 20V d.c. to the SQ decoder, 18V d.c. to the a.m. tuner, f.m. tuner head transistors, i.f. stages and the stereo multiplex decoder. The values of the ladder are chosen so that either the a.m. or f.m. tuner may be on, or both tuners off, without disturbing the voltages of the other supplies significantly.

The components C26 and C27 and resistor R37, together with D18 are associated with the tuner unit (yet to be described) and in fact appear on that unit's board. They are shown here merely for completeness.

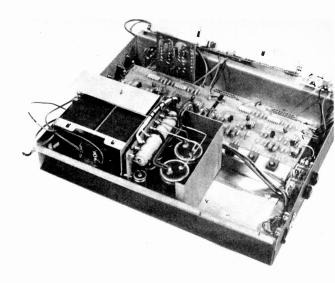
CONSTRUCTION

The power supply p.c.b. master negative is shown in Fig. 3.2 which indicates the simplicity of the circuit in practice.

The component layout appears in Fig. 3.3 which is self explanatory. Capacitors C19 and C20 are too large for board mounting and are in fact clamped in the main chassis.

MAIN CHASSIS

The main chassis member is a rectangular boxform which in fact sits on its back in the final assembly. Thus it is used as a trough with components mounted within it rather than in the more conventional manner.



General view of the "trough" form of the main chassis for the Rondo showing the orientation of the power amplifiers and power supply with some of the general wiring in place. Plug and socket holes are not sized as this will depend on items selected

The chassis is shown in the bent-up form in Fig. 3.4. The material used for the prototype was 20 s.w.g. mild steel plate but of course this may present some difficulties to constructors since it is not easy to bend. Thus it is possible to use 16 s.w.g. "half-hard" aluminium sheet which is somewhat easier to bend or, indeed, to fabricate the structure from flat sheets cut to size. For this reason only the major dimensions have been given as it is felt that each constructor will probably adopt his own style of construction.

It should be added that the chassis, in the prototype design, contains all the various parts of the system and supports the outer wooden case and facia which is in the form of a sleeve which slips over the unit and is fastened to the chassis.

PLUGS AND SOCKETS

As individual constructors may wish to vary the types and sizes of plugs and sockets used the positioning shown is open to variation to suit though of course major shifting of components is not advised as this can create feedback paths for which the equipment is not designed.

CHASSIS SCREENS

The power supply includes the mains transformer, board and main smoothing capacitors and the whole is placed behind a shield to cut down hum pick-up at the pre-amplifier and elsewhere.

Two screens are used, as illustrated in Fig. 3.5 and 6. The smaller acts as a support for the power supply p.c.b. and is fastened to the transformer using 4BA screws. The p.c.b. is then attached to the screen using 6BA screws and $\frac{3}{8}$ in spacers.

The larger screen is assembled later in the sequence and is held in place using self-tapping screws.

Note that the transistors in the power amplifier last month should be type MJE3055 \mathbf{K}

Next month we will finalise the mechanical assembly details and discuss the unit interwiring





DOGGED RESISTANCE

Honestly, the trouble to which people go in order to maintain their various equilibria.

The latest in this mania for status quo manipulation appeared in a letter to a professional U.S. magazine. This, in fact, took the form of a reply to an earlier piece of correspondence which complained about noise from a barking dog, and which simply requested "Help! Widget wanted to drive the Mutts"

Help, if this is what it could be referred to, appeared a month later as a package deal comprising a 20kHz oscillator and (twenty watt!) amplifier combo driving a hi-fitweeter horn. The suggestion was, that by continuous use, although no human would be able to hear the thing, poor ole' Fido would make such an awful din that his ever-loving owner, thinking he'd gone out of his tiny mind, would be just itching the next day to have him put down!

Now no one could accuse me of holding any torches for the canine community, nevertheless, I cannot help feeling that this is a singularly pernicious use of electronics.—One couldn't help noticing that the correspondent had asked to have his name withheld!

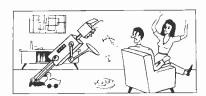
LSI or AU PAIR?

The number of times we have to hear about the feasibility of robot "flunkies" in the home, and how extraordinarily easy it's all going to be once they're installed. All you need is a handful of super-LSI chips, and you'll never look back! No doubt, the electronics represent no

insurmountable problem, but how practical would something like this be, even assuming the mechanical aspects could be solved?

The variety of jobs that can be performed by automated equipment is already overwhelming; what with tumble driers, telephone answerers, record players, etc., it might even be questionable whether a place exists for robots in the true sense of the word. Yet, I read, just last week, that Dr George Müeller of Systems Development Corporation, California, reckons that a computerised set-up of this sort would be a viable proposition within, say, ten years.

Once written, I suppose, executive programmes having the same general format could be hard-wired into the devices prior to despatch. The "crunch" might come when the new owner needed to programme the confounded thing for the sundry tasks around the home.



Imagine trying to write a programme for accomplishing the "simple" job of pouring drinks. You would need to tell it to "come out of your storage cupboard", "turn left after three feet", "continue for nine feet seven inches" (at this point you might be wondering whether it only understood metric), "stop", "move right arm slowly towards cupboard right-hand handle", "grip handle gently and rotate it 180 degrees counter-clockwise" (supposing that the catch didn't release until 181 degrees?).

"Carefully open door 120 degrees, then release handle", "lower arm, and scan inside cupboard for port bottle (does robot think I mean the bottle on the left?), "if located, remove the cork in this bottle and pour a quantity of liquid from the bottle into each of the glasses situated below the bottle rack", "ensure that each glass is only filled to the \frac{3}{4} mark", etc., etc.

With such a system, one need only forget about junior's toy duck, 'plum in the middle' of the automaton's path, to give it a full-blown seizure!

A machine able to rely on learning techniques might prove to be a better proposition, always provided you could tolerate the broken glass, wine-sodden carpets, and the bill from the psychiatrist while it got the hang of things!

MAINSTREAM

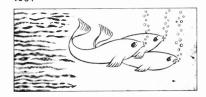
By nature (I'm told) the trout is a lazy sort of fellah; loves to bask himself in the sun, and only makes a sudden dart across the current if, in his myopic way, he fancies he's seen a fly somewhere on the surface. In this life, though, there is rarely anything that escapes exploitation and the trout represents a typical case in point.

Recently, you see, the waterworks at Boran sur Oise, near Paris, have overcome some difficulties associated with the monitoring of pollution levels by recruiting the aid of this delicious member of the aquaticae. Seemingly, ordinary detection/warning systems in this application are insufficiently sensitive; as a corollary, the waterboard's laboratory have "roped-in" three good-sized brown trout to do the job instead.

Each animal is located in a glasssided tank fed from normal supply water and running at a rate consistent with typical river current. Normally, the trout orientates himself up-stream; however, if conditions cause the water to be slightly polluted, he'll about-turn to face downstream.

By fitting up each of the fish with tiny electrodes, an alarm can be triggered whenever all three fish do a simultaneous "flip round". Individuals turning-tail are ignored by logic ANDing circuits to avoid false alarms.

Several days of this treatment do not appear to upset these creatures, in spite of the fact that if any of them stop swimming they are given an electric shock into the bargain. Since these animals are required to work in a sterile environment one wonders what technical arrangements have been made to feed 'em too!



BACK NUMBERS WANTED

Anyone who can supply the undermentioned are asked to communicate directly with the reader.

December 1971 Mr. J. Bickerstaffe, 15 Deanshill Close, Stafford.

November 1969, February to April 1970 Mr. S. R. Dunning, 103 Glebe Road, Deanshanger, Wolverton, Bucks.

July to October 1972 Mr. J. Y. T. Lee, 20 Wong Chuk Hand Road, 2nd Floor, Aberdeen, Hong Kong. September 1972 Mr. K. R. Whitbread, 14 The Orchard, Bedford Park, London, W4 IJX.

Any Back Numbers # Drive, Rodborough, Stroud, Glos.

May 1972
Mr. J. Hoppe, Murraymead, Bracknells Lane, Hartley Wintney, Basingstoke, Hants.

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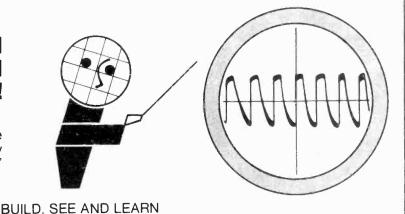
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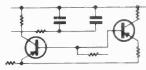
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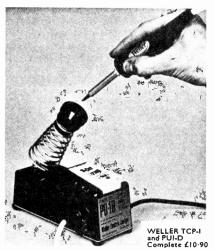


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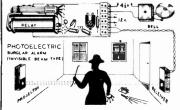
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LOGIC TUTOR

EXPERIMENTS



by M. J. Hughes

THE circuit we left you to puzzle over last month was another form of the EXCLUSIVE OR gate. The Boolean expressions for the various nodes are shown in Fig. 1. The outputs of the two gates that are WIRED OR'd together would have been $\overline{A} + \overline{B}$ and A + B respectively without the link between them but as soon as the link is made these two functions become ANDed together and then it is a simple matter of Boolean manipulation to show that the output function is EXCLUSIVE OR.

Binary Addition

It is necessary to understand the basic principle of adding together two binary integers. The rules are very simple—in fact exactly the same as in conventional denary arithmetic except that whenever you get a sum greater than I you must carry over a digit into the next column. Let's take the simple case of adding two single digit numbers together. We'll call the digits A and B to differentiate between them and show the sum and carry—when necessary—in the following simple table. Note that we are carrying out pure addition here and in this instance + means plus and not OR

In this list of sums we have used every permutation of the two numbers. Compare the arithmetic with the truth table for the EXCLUSIVE OR gate shown in Fig. 2. If we used electrical signals to represent the numbers we wanted to add together and could accept an electrical signal as an answer you can see that the output of an EXCLUSIVE OR gate gives use a true representation of the sum of the two binary integers. It does not, however, give us the carry digit when we want to add I and I. In addition we only need a carry when we have I AND I so it is a simple matter to provide this output from the same pair of inputs by introducing an AND function.

Fig. 3 shows how this can be done by taking the \overline{AB} function—generated at the centre node of the EXCLUSIVE OR—and inverting it. You can see that the truth table for the circuit shown in Fig. 3 is an exact replica of the answers we would wish to get when carrying out a binary addition. The circuit is called a half adder. As is often the case there are various ways of designing half adders—you now know at least three ways of making the EXCLUSIVE OR function so try making some more half adders yourself on the Logic Tutor.

You might query why this is called a half adder. The reason is that when we come to add together two multidigit numbers (see Fig. 4) our current circuit is quite capable of dealing with the least significant column (i) but when we come on to the other columns we have to be able to add together the respective digits of numbers A and B but also have to be able to add in any carry over that was generated in the previous (lesser significance) column. Thus to add multidigit numbers together we must have a circuit that can handle three inputs.

A circuit that will do this is called the full adder and we shall deal with this next month.

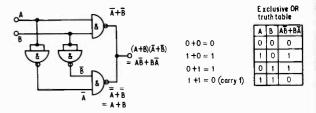


Fig. 1. Answer to last month's problem. The circuit is an EXCLUSIVE OR

Fig. 2. Comparison of binary addition with EXCLUSIVE OR truth table

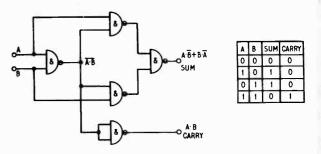


Fig. 3. A half adder is an EXCLUSIVE OR plus an AND function of the inputs

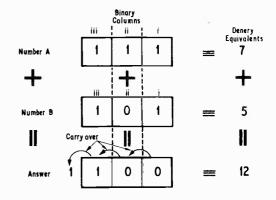


Fig. 4. When carrying out addition of two multidigit numbers a half adder is capable of dealing with column (i) but when it comes to column (ii) there must be provision for three inputs, digit (ii) of A + digit (ii) of B + possible carry over from column (i). The same applies to all higher power columns

PE Sound Synthesiser 10

KEYBOARD

With these added units the keyboard becomes a unique musical instrument

When the Sound Synthesiser was originally presented for publication it was the intention that it should be classed as a general purpose instrument which could be exploited in the widest possible number of ways and yet retain a basic simplicity of design and ease of construction. During the course of the series however many readers have commented that the musical capabilities of the instrument have been severely restricted by the lack of logarithmic v.c.o.s and thus the oscillator to be described in this article has been included in the keyboard unit in the hope that it will put matters right.

Prospective constructors should note that although the logarithmic v.c.o. is based on the linear v.c.o. which was described in Part 3, the parameters of operation are quite different and the setting-up is rather more critical if it is desired to operate the device within the limitations of a precise range of

control voltages.

Although limited to about 11 octaves range with the design values given, it is possible to adjust the operating points to cover a much wider bandwidth. The prototype has operated quite happily from less than 1Hz to greater than 150kHz with a control voltage swing of about 11V and there is no reason to suppose that it could not reach 1MHz, or greater, if the integrator output voltage were to be reduced and a faster comparator employed. As a result it is possible that the oscillator may be suitable for a number of applications outside the sphere of sound synthesis.

The v.c.o. is shown in block schematic form in Fig. 10.1. The control voltage to the oscillator is modified in a differential input summer and then applied to a constant current generator housed in a "transistor oven". The output of the current generator is then led to an integrator/comparator stage through the medium of a current switch which is controlled jointly by the comparator and inverter. The triangular wave output is led to a waveform shaping circuit which provides a sine wave having a very low harmonic content.

DESIGN CONSIDERATIONS

It was decided that the best way of providing a log law performance to the basic v.c.o. outlined in Part 3 of the series was to utilise the entirely predictable logarithmic relationship between the base/emitter voltage (V_{be}) and collector current (I_c) of a bipolar transistor.



• Start Here ! .

This month we start detailing the operation and construction of the keyboard unit for the synthesiser. Regular readers of P.E. who may have been put off by the apparent complexity and/or cost of the synthesiser project as a whole may be interested to learn that the facilities offered by the keyboard unit are such as to allow the instrument to be classed as a music synthesiser in its own right, and at an overall cost of less than £80. Main features of the instrument are as follows:

Two tracking oscillators featuring a variable logarithmic law which allows compatibility with a wide range of control voltages and provides square, triangular and high purity sine wave outputs.

A novel "floating" divider system which greatly simplifies tuning the instrument and by means of which the compass of the keyboard can be swung, in tune, from a low frequency of about 6Hz to a high frequency greater than 27kHz. The divider features a switchable "span" facility.

Two modulation amplifiers by means of which the oscillators may frequency modulate one another either separately or at the same time.

Two analogue memory circuits which retain the last programmed divider voltage to either oscillator. Separate portamento controls are incorporated in the hold circuits giving six values of delay from instantaneous to one second.

Two simplified envelope shapers incorporating voltage controlled amplifiers each having a variable Attack and Decay characteristic and featuring a switched percussive attack.

A simple two channel, fixed gain mixer.

Finally, the keyboard divider system incorporates a link switch which allows the two oscillators to be programmed independently by the lower 18 and upper 31 keys respectively. This particular feature greatly improves the "live performance" possibilities. An independent p.s.u. is included.

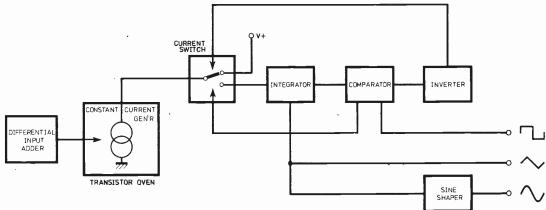


Fig. 10.1. Block schematic of logarithmic v.c.o.

Although the current generator could be almost any discrete transistor the effect of ambient temperature variations would be such as to cause the current demand to vary quite widely and thus adversely affect the "tune" of the oscillator. Consequently it was decided to make use of the inherently close thermal and electrical matching between transistors mounted in a monolithic integrated circuit — the ML3046P and CA3046 both having been tried in the prototype. Extremely close thermal stability is ensured by employing two of the transistors in the array as a heater and sensor respectively and terming the whole arrangement as a "transistor oven."

Fig. 10.2a shows a detailed arrangement of the "oven" while Fig. 10.2b shows the pin connections for the 3046 device. Fig. 10.3 illustrates how the V_{be} of transistors on the array will vary over a wide range of temperatures and provides a guide as to the actual temperature of the chip when the V_{be} of the sensing transistor is known.

In Fig. 10.2a Q1 serves as the heating element while Q2 is used to sense the temperature of the chip. The $V_{\rm be}$ of Q2 is compared with a reference

voltage set up by R3, VR1 and R4. The reference voltage will correspond to a temperature which is considerably higher than ambient and will thus be at a lower value than Q2V_{be} when power is first applied. Thus the comparator will switch positive and turn on Q1. As the temperature of the chip rises Q2V_{be} will fall to a point where it is equal to or less than the reference voltage at which time the comparator will switch negative and turn off Q1.

The criteria determining the value of R1 are as follows:—

- 1. The temperature of the chip must be set considerably above normal ambient conditions (say 40 degrees to 45 degrees Centigrade).
- 2. The combined power dissipation of Q1 and Q2 must exceed, by a wide margin, the combined power dissipation of the remaining transistors in the array.
- 3. The current switching of Q1 must not be so violent as to impart a significant jitter to the oscillator waveform.

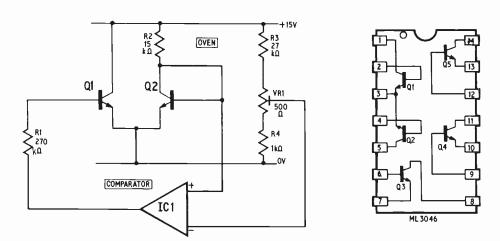


Fig. 10.2(a). Detailed arrangement of the transistor oven; (b) pin connections of the ML3046 and CA3046. Note that pin 13 must be connected to the most negative point in circuit

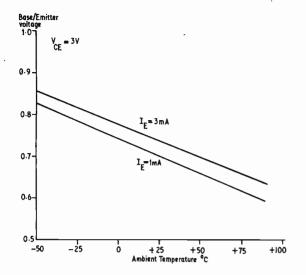


Fig. 10.3. Showing the change in $V_{\rm \,be}$ for variations in ambient temperature for the ML3046 and CA3046

In practice the value of R1 which gives the closest approach to the above criteria is 270 kilohms. With this value it is possible to set Q2V_{be} to 680mV, that is approximately 45 degrees Centigrade while the combined dissipation of Q1 and Q2 is about 90mW as opposed to the combined dissipation of Q4 and Q5 (the current generators) of about 1.4mW. Detailed setting-up instructions for the "oven" are given later in this article.

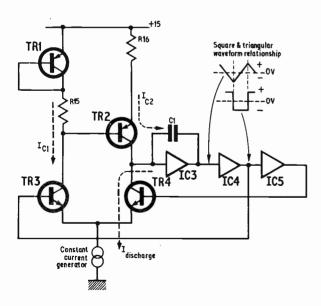


Fig. 10.4. Simplified layout of oscillator section showing operation of current switch

CURRENT SWITCH

Having thus established a thermally stable current generator it now remains to couple this to the input of the v.c.o. In the linear version of the v.c.o. a diode bridge was employed which was switched entirely by the action of the comparator. While this system could no doubt still operate satisfactorily a rather more sophisticated version has been adopted which utilises transistors in place of diodes. The so-called current switch is illustrated in Fig. 10.4. Assuming that the output of IC4 is positive to start with, the operation of the current switch is as follows.

With IC4 positive, the output of IC5 is negative and thus TR3 and TR4 are on and off respectively. With TR3 on the current generator demand will cause a drain across R15 and thus make the base of TR1 more negative than its emitter. TR1 will then turn on in proportion to the demand of the current generator and a current flow I_{c_1} is established through TR1, R15 and TR3. I_{c_1} will set up a p.d. across R15 which will have the effect of biasing on TR2 and establishing a second current flow, I_{c_2} through R16 and TR2.

This latter current flow is entirely dependent upon the demand of the current generator which is varying the p.d. across R15 and thus TR2 may be said to track the demand of the current generator. The closer the matching of R15/R16 and TR1 and TR2, the better will be the tracking and the better the symmetry of the integrator output waveform. Thermal stability is not a problem with TR1 and TR2 since the arrangement gives an equal and opposite reaction between these transistors for any variation in temperature.

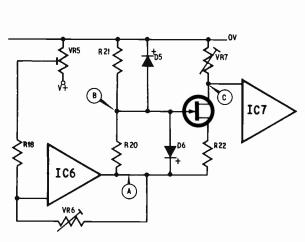
 $I_{\rm c2}$ causes the integrator to ramp negatively and at a predetermined negative level IC4 will switch negative and IC5 positive. Thus TR3 and TR4 are now off and on respectively and since TR3 is off so also is TR2 and the current generator now discharges C1 via TR4 at a constant rate. The integrator will thus ramp positively until the switching point of IC4 is reached at which time the cycle repeats.

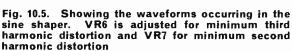
SINE WAVE SHAPER

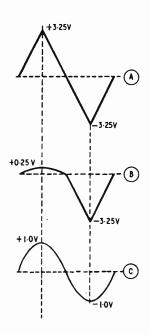
The sine wave shaper used in the v.c.o. has been adapted from a design by D. T. Smith which appeared in *Wireless World* (Feb. 1973). Fig. 10.5 shows a detail of the circuit together with the waveforms presented at various points when the circuit is producing a sine wave having a low harmonic content.

The sine-shaper utilises the non-linear characteristics of a field-effect transistor in order to produce the desired output waveform and thus the operating points are quite critical. R20, 21 and D5, 6 provide a network which allows the gate to track the input signal and apply the necessary degree of "pinchoff" as the source signal approaches its maximum on both positive and negative half-cycles. Since the gain of the f.e.t. is changing continuously relative to the input signal the device may be considered to be operating in the ohmic region as a voltage variable resistor.

In Fig. 10.5 VR5 controls the d.c. offset of the input waveform and it is necessary to compensate for slight asymmetries which could be introduced in







the oscillator section. Over adjustment of VR5 in either direction can provide an alternative output waveform rich in harmonics.

VR6 sets the input signal level to the network. Too low a setting and the output waveform will be triangular while too much gain will give a waveform which is virtually a rounded off trapezoid. Careful adjustment of VR6 enables third harmonic distortion on the output to be reduced to a minimum.

VR7 controls the output amplitude of the signal. Too great a setting and the sine wave will become peaky while too low a setting will result in a squat, flattened out, waveform. Thus VR7 may be considered to control the second harmonic level and should be adjusted to minimise this characteristic.

BUILDING THE OSCILLATORS

The theoretical circuit of the complete oscillator module is shown in Fig. 10.6 while the recommended circuit board layout is shown in Fig. 10.7. It is strongly urged that construction of the module follows the guidelines suggested otherwise the interdependance of some of the controls is likely to make the final setting-up something of a nightmare.

Construction should start with assembly of the transistor oven, differential input summers and current generators. R1 should be 470 kilohm and left with fairly long leads to facilitate exchange during the final setting-up. Position the temperature control, preset VR1 to the R3 end of its travel to ensure that the comparator goes negative and thus turns off Q1 when power is applied. R10 should then be temporarily linked to the 0V rail and a decision taken as to the fate of R11. This resistor is optional in that it may be utilised to provide a third controlling input to the v.c.o. or it may be omitted at this stage without affecting anything. However, if it is decided to include R11 for the purpose of possible future additions to the circuitry it is important to remember that its presence could possibly compromise the setting-up of the oscillator if certain precautions are not taken.

If a third programming signal is to arrive at the oscillator from a low impedance source such as the output of another operational amplifier it is important that R11 also be temporarily linked to the 0V rail. Alternatively R11 may be left open circuit with the idea of using it only for the provision of occasional programming signals during which the tuning of the oscillator will have to be adjusted by means of VR4. In the setting-up instructions which follow R11 is open circuit.

SETTING UP

Since only one current generator can be set-up at a time the base of the second one (Q5) should be temporarily linked to the -15V rail to reduce the possibility of accidents. Set VR3 so that its value is 530 ohms. Note that if the positive lead of the ohmmeter is coupled to the -15V rail for this measurement the reading will not be compromised by the forward conduction of Q4. Set VR2 to the R6 end of its travel to ensure the minimum forward bias of Q4.

Set VR4 to its minimum setting and connect a milliammeter between the collector of Q4 and the OV rail as shown in Fig. 10.8. As a safety precaution set the milliammeter to the 25mA range to start with. If the wiring up has been correctly carried out the milliammeter will show no reading when power is applied. On the other hand if the base of Q4 is incorrectly biased towards the positive region Q4 will pass a large current on switch-on which, besides possibly destroying the transistor, could also damage the meter.

Apply power and if no meter movement is observed reset to the 1mA range. Again there should be no obvious reading. Adjust VR2 towards the R5 end until the pointer of the meter makes a definite upward movement. At this point swing VR4 carefully over its full range and observe the maximum current drawn through Q4. The actual figure can vary quite widely at this stage and will probably be in the region 0.25-0.75mA.

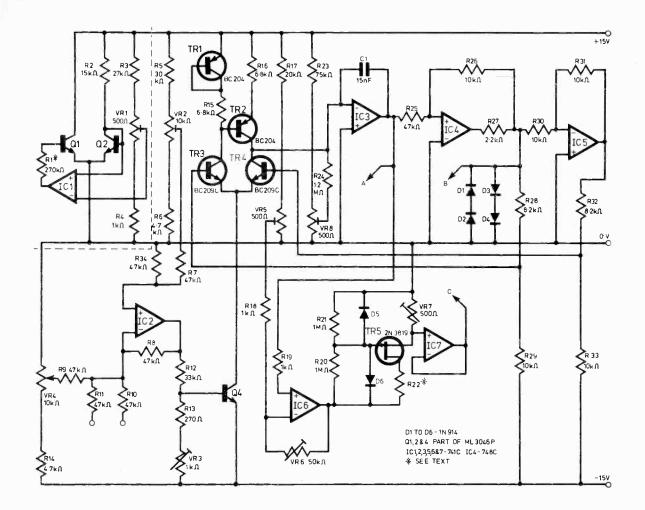


Fig. 10.6. Circuit of complete oscillator module. Points A, B and C are coupled to waveform selector switch

Set VR4 so that the meter is reading 0.2mA and place a slightly moistened finger on IC8—the transistor array. The meter reading should be seen to increase quite rapidly. Having made the above adjustments switch off and remove the meter from the collector circuit of Q4. Without disturbing the setting of any of the potentiometers temporarily link Q4 collector of the 0V rail. The same adjustments should now be made to the circuitry around Q5—having first removed the base/negative rail link.

In this case, however, leave the meter in circuit for the next stage of setting-up which entails fixing the thermal working points of the transistor oven.

Couple the oscilloscope to the output of the comparator IC1 which, at this point, will be about -14V. Carefully adjust VR1 until the comparator switches to +14V. At this time two things will start to happen. The first is that the meter will show a progressive increase in reading which, if the adjustment of VR1 has been made with care, should not exceed ImA. The second is that within a second or so of the comparator switching positive the oscilloscope trace, which starts as a straight line, will begin to display a varying waveform.

At first the oscillation will zero about a point

approximately 12V positive, but, as the chip temperature stabilises, the zero point will move down to about +6V to +9V. Monitor the base emitter voltage of Q2 which should be in the region 650mV to 680mV.

LOW LOAD POINT

At this time all four connected transistors are contributing towards the heating on the chip. QI with its 470 kilohms base resistor will be passing about 3mA and contributing about 45mW, Q2 in passing 1mA will be contributing about 15mW whilst Q4 and Q5 will be contributing about 6mW jointly. These latter devices, however, are passing considerably more current than they will be required to do when coupled to their respective oscillators and running at high frequency. It is therefore necessary to establish a low load set point to ensure that the oven maintains the same temperature over the full working range of the oscillators.

On both Q4 and Q5 reduce VR4 to its setting and readjust VR2 hard over towards R6—in other words reduce the forward bias on the transistors to the absolute minimum. These adjustments will cause

VOLTAGE CONTROLLED **OSCILLATORS**

COMPONENTS.

LOGARITHMIC V.C.O.s. (2 required)

Resistors

270k Ω (see text) R1*

R2* $15k\Omega$

R3* $27k\Omega$

R4* $1k\Omega$

R5

 $30 k\Omega$ 2% metal oxide $4.7 k\Omega$ 2% metal oxide

R7-R11 47k Ω (5 off) 2% metal oxide

R12 33k Ω 2% metal oxide R13 270 Ω 2% metal oxide R14 4.7k Ω 2% metal oxide

R15-R16 6-8kΩ 2% metal oxide

R17 $20k\Omega$

R18-R19 1k Ω (2 off)

R20-R21 $1M\Omega$ (2 off)

R22 See text

 $75k\Omega$ R23

 $1.2M\Omega$ R24

R25

47k Ω 2% metal oxide 75k Ω 2% metal oxide R26

R27 2.2kΩ

R28 8-2kΩ

R29-R31 10kΩ (3 off)

R32 8-2kΩ

R33 $10k\Omega$

47kΩ 2% metal oxide R34

Capacitor

C1 15nF

Potentiometers

VR1* 500 Ω cermet preset

VR2 10kΩ cermet preset

VR₃ 1kΩ cermet preset

10kΩ midget moulded VR4

carbon

VR5 500Ω cermet preset

 $50k\Omega$ cermet preset VR6

VR7 500Ω cermet preset (see

text)

VR8 500Ω cermet preset

Integrated Circuits

IC2, IC3, IC5, IC6, IC7 741C (5 off)

IC8* ML3046P (Q1, Q2, Q4, Q5)

IC4 748C

Transistors

TR1-TR2 BC204 (2 off)

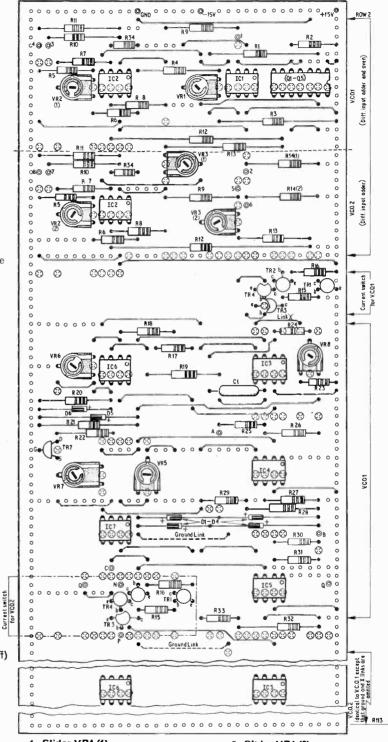
TR3-TR4 BC209C (2 off)

TR5 2N3819

Diodes

D1-D6 IN914 (6 off)

Note: For components marked with an asterisk only 1 off is required.



1. Slider VR4 (1) 2. Hi-end VR4 (1)

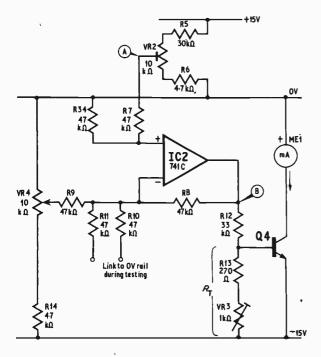
3. R10 Control Input (1) 4. Spare Control Input (1) 5. Slider VR4 (2) 6. Hi-end VR4 (2)

7. R10 Control Input (2) 8. Spare Control Input (2)

Fig. 10.7. Board layout of oscillator module. Link point Q to Q. Link point N to junction of R33 and R33 or V.C.O.2. Link point P to junction of R28 and R29 on V.C.O.2

the oven temperature to fall and thus the comparator switching waveform will tend to move more positive and revert to a straight line trace.

Monitor the base/emitter voltage of Q2 again which, with the comparator positive, should not be greater than 680mV. If it is, or if, after a few seconds, the comparator waveform does not show signs of rippling, then the heat dissipation of Q1 is insufficient to maintain the set temperature. Much,



 $V_{BE} Qz = 680 mV: R_T = 750 \Omega$

	4 BE 65 - 00011	4. KI - 100	
Bias Voltage "A"	Total Voltage "B" Bias + Control	lc	V _{BE} (Q4) (mV)
2.74	2.74	12nA (calc)	395
2.74	5.75	160n A	461
2.74	7.76	600n A	500
2.74	8.76	1·5μA	528
2.74	9.76	3·7μA	550
2.74	10.78	7·0µA	570
2.74	11.78	20·0µA	595
2.74	12.78	41·0μA	617
2.74	14-39	60·0μA	630
	Referred to 0V Rail		Referred to -15V

Fig. 10.8. Measurements of Q4V $_{be}$ made with Avo 8 which illustrate the degree of error that is possible when interpreting the reading

Rail

of course, depends upon the ambient conditions when these measurements are being made and it is best to set up the low load point at the coolest temperature at which the v.c.o. is likely to be operated.

Under cool conditions therefore, if the comparator still does not ripple, it will be necessary to adjust the value of R1 to increase the current through Q1. In the prototype a value of 270 kilohms proved to be satisfactory and gave the desired comparator switching waveform at low load. The value of R1 is quite important because if it is too low the current switched by Q1 will be excessive and cause a considerable degree of jitter on the oscillator waveform.

Monitor $Q2V_{be}$ once more and adjust VR1 as necessary to bring the voltage to 680mV. All setting-up on the prototype oscillator was carried out at this value and it is important to bear in mind that the values of all subsequent measurements bear a close relation to this figure. If $Q2V_{be}$ is lower than 680mV this implies that the temperature of the oven is higher and thus Q4 will pass a higher current for any given value of control voltage. Since the frequency of the oscillator is linearly related to the current through Q4 then the frequency will also be higher.

The reverse will occur if the value of Q2V_{be} is higher than the specified figure. However, for small variations from the specified value, the performance of the oscillator will still remain wholly logarithmic and sufficient tolerance has been allowed in the biasing control VR2 to compensate for such variations. Once satisfied that the upper and lower temperature set points have been correctly established VR1 may be locked with a small dab of non-conductive adhesive such as Araldite. Link the bases of Q4 and Q5 temporarily to the -15V rail.

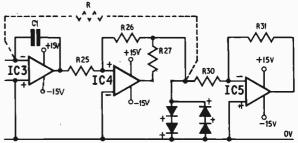


Fig. 10.9. Basic circuit elements of the oscillator and inverter (IC5). Resistor R is any value between $2k\Omega$ and 1M Ω and is temporarily coupled as shown to prove oscillator function

PROVING THE OSCILLATORS

Fig. 10.9 shows the circuit elements involved in the construction of the oscillator section. When these items have been assembled temporarily connect a resistor of between 2 kilohm' and 1 megohm from the junction of R27/R30 and the inverting input of IC3. With an oscilloscope monitoring the output of IC3 apply power and a triangular waveform will be observed at a frequency dependent upon the value of linking resistor chosen. Having proved the functioning of the oscillator assemble one current switch and couple to a current generator having first removed its base shorting link. Resistors R28/29 and R32/33 should be added at this time. The purpose of these resistors is to establish a negative bias point for transistors TR3 and TR4 in the

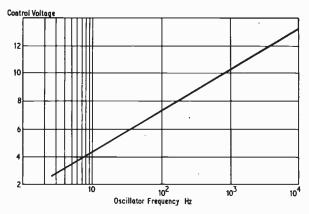


Fig. 10.10. Control voltage/frequency relationship

current switch and whereas the exact value of the resistors concerned is not entirely critical it is important that R28/32 and R29/33 be as closely matched as possible to ensure that the bias points of TR3 and TR4 are the same.

A quick check may now be made to ensure that the separate circuits work together as a complete unit. Remember to reset VR2 and VR4 to about mid position. Depending upon the setting of VR3 it is possible that rotation of VR4 will cause the oscillator to go into saturation towards either extremity but this is not important at this stage and

will be dealt with during final setting up.

The next stage is to build the second oscillator following the same general pattern and, having established that it functions, begin the process of matching the performance. The closest possible matching of performance will be obtained if relatively close tolerance components have been used in the construction and, for this reason, 2 per cent resistors have been specified throughout. In particular it is prudent to obtain a matched pair of integrating capacitors (C1).

BIASING

Firstly it is necessary to establish a value of minimum bias on Q4 and Q5 which will support oscillation. Due to the inherently close electrical matching of the transistors on the 3046 the same level of bias will result in current flows through the transistors which are, for all practical purposes, identical. In the prototype it was found that the output of IC2 was at +2.74V referred to the OV rail for an oscillation frequency of 0.2Hz.

Set VR4 to its minimum position and adjust VR2 until the output voltage on IC2 (both oscillators) reaches +2.74V as above. Monitoring the integrator output waveform on both oscillators at the same time, if possible, adjust VR3 on Q4 and Q5 with the greatest care so that both oscillators are running at the same frequency—the exact rate is not critical. With VR4 still at its minimum setting apply an external control voltage to R10 on both oscillators at the same time having first broken the temporary link connecting R10 to the OV rail. A fresh 9V battery with a suitable potentiometer coupled across its terminals is ideal for the purpose of providing the control voltage.

Connect the positive end of the battery to the OV rail and the slider of the potentiometer to both R10's. With the slider hard negative monitor the

output frequencies of both oscillators, preferably at the same time, and confirm that they are running at the same frequency which should be in the region of 3kHz. Note that since the oscillators are not phase locked a certain degree of drift between them is almost inevitable and it should be the aim to reduce the amount of drift, by adjustment of VR3, to within 1 per cent or better of the frequency being monitored. Thus for a frequency of 3kHz a drift of about 30Hz would be at the limits of acceptability. Compare frequencies at various settings of the potentiometer to ensure that the frequencies and drift relationships remain stable over the full range.

OFFSETTING SATURATION

If, at the minimum setting, either or both oscillators go into saturation due to the adjustments made to VR3 it will be necessary to establish a slightly higher bias point by re-adjustment of VR2 and then repeat the whole of the setting up procedure so far outlined. Careful adjustment will result in a pair of oscillators which track the control voltages very accurately. It should not be the aim to reduce the drift between the oscillators to a very low figure as the beat frequency introduced by a drift of 0.5 per cent to 1 per cent will add interest and colour to the sound when both oscillators are being programmed in harmony. On the other hand a very low beat frequency can add quite unpleasant characteristics to a sound.

When satisfied that the oscillators are tracking over the full range of the control potentiometer the control voltage measured at the output of IC2 should be plotted against the frequency of oscillation at various points in the range. Fig. 10.10 shows the result of plotting the performance of the proto-

type oscillators.

F.E.T. CHARACTERISTICS

Before beginning the assembly of the sine shaper it is necessary to determine the exact characteristics of the f.e.t. which is to be used in the circuit. The two operational parameters which require to be known are the saturation current (I_{DSS}) at zero gate bias and the gate bias required to reduce the current through the device to negligible proportions. Fig. 10.11a/b illustrates the methods of making the measurements specified. If a variable voltage source

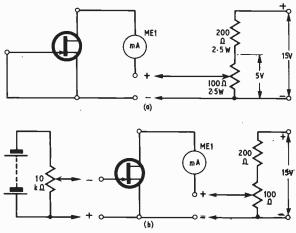


Fig. 10.11(a). Method for determining $I_{\rm DSS}$ at $V_{\rm GS}{=}$ O(b) method for determining $V_{\rm GS}$ when $I_{\rm DS}{=}$ O

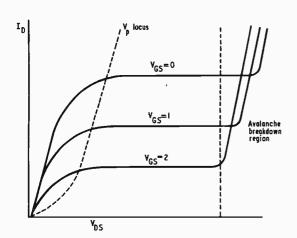


Fig. 10.12. Graph showing how $V_{\rm p}$ and $I_{\rm DSS}$ vary with $V_{\rm GS}$

is available it is preferable to use this in the source drain circuit rather than the divider arrangement illustrated and thereby gain the benefit of greater accuracy in the measurements.

The first stage is to measure the $I_{\rm DSS}$ at zero gate bias. With the variable voltage source at zero volts and the milliammeter on the ImA range gradually increase the voltage setting, plotting, at various

stages, the current/voltage relationship. The point which is of interest is that at which further increases in voltage result in only a very small increase in current through the device. The voltage at which this phenomenon first occurs is known as the pinch-off voltage (V_p) and should be carefully recorded. After the pinch-off point has been reached the voltage may be increased quite significantly with very little increase in current until the avalanche breakdown region is reached. At this point the current through the f.e.t. will increase hugely and almost instantaneously and result in the destruction of the device. Hence the requirement to plot the measurements very carefully and note the point at which V_p is reached.

Fig. 10.12 shows a typical family of curves for any one f.e.t. and depicts the way in which the pinchoff voltage reduces as the gate bias is made progressively more negative with respect to the source. The next measurement to make therefore is the point at which the current through the f.e.t. reduces to negligible proportions and the set-up for doing this is illustrated in Fig. 10.11b. Having made the connections shown and with the 10 kilohm potentiometer wiper at the positive end of its travel adjust the variable voltage source so that the meter is indicating the I_{DSS}. At this point gradually advance the wiper and note that the meter reading reduces in proportion. Switch to a lower range on the meter as required and advance until the reading is zero. At this point carefully measure and note the voltage across the gate/source of the f.e.t.

The value of resistor R22 in the sine shaper is calculated on the basis of the readings above in the following way:—

$$R22 = \frac{V_p}{2.I_{DSS}}$$

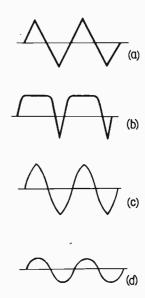


Fig. 10.13. Showing the effect on the sine shaper output waveform of adjustment of VR6 and VR7(a), VR6 too low (b), VR6 too high (c), VR7 too high (d), VR7 too low

The $V_{\rm p}$ in the above calculation refers to the value of gate/source voltage at which the current through the f.e.t. is zero. The value of the resistor will depend on the actual characteristics of individual f.e.t.s but would normally be expected to be quite small. In the prototype, for example, R22 was 180 ohms in one shaper and 270 ohms in the other. The value of VR7 should be chosen to provide a fairly wide margin of adjustment over the calculated value of R22 and in most cases a 500 ohm preset would be satisfactory.

ASSEMBLY

Having completed the above measurements the sine shaper can now be assembled. Bear in mind that f.e.t.s can be rather tricky to handle and it is a wise precaution to solder all the other components in position before actually inserting and fixing the f.e.t.

Setting-up the sine shaper consists of adjusting the values of VR5, 6 and 7 to provide the optimum sine wave. With power on adjust VR4 so that the oscillator is running at about 3kHz and monitor the output of IC7. The preset adjustments should be made with reference to Fig. 10.13 which illustrates the various waveform characteristics associated with these controls.

If a sine-wave oscillator is available it is helpful to compare the output of the shaper with a "genuine" sine wave of the same frequency. The scale of the waveform on the oscilloscope screen should be as large as possible for this purpose. This latter procedure was carried out with the prototype shapers and resulted in a sine wave output having a total harmonic distortion of only 1 per cent. A wave analyser or distortion meter if available could enable a higher purity sine wave to be obtained.

Next month: Envelope shaper, mixer networks and analogue memories for the keyboard unit.

QUICKIES



4 EASY-TO-BUILD PROJECTS FOR THE WINTER EVENINGS

THERMON FIFE



A temperature measuring device with many applications

HENRYS RADIO 2/9/74.

THE electronic thermometer has been proposed in various forms, usually associated with a thermistor detecting element. This latter is a resistor the value of which changes with changes in temperature, hence the name from thermal resistor.

However, whilst capable of providing an indication with fairly simple circuitry, this device is basically non-linear, the change in resistance for a given change in

temperature is not the same for all temperatures.

There is a much smaller and, in linearity terms, more accurate device readily available, the forward - biased semiconductor diode.

Diode Probe

It is admitted that the range over which a small silicon switching diode can be used in temperature terms is limited to the area which will not damage the device, namely from around -60°C to $+180^{\circ}\text{C}$. In addition, the variation of resistance and thus voltage across the diode is very small for a change in temperature of 1 degree C, in the region of 2.5 to 3.5mV.

Thus the change needs to be amplified in some way if it is to be useful.

The advent of cheap integrated circuit amplifiers has provided the answer to that problem and the

operating range of these diodes is quite sufficient for most applications.

Circuit

· A number of alternative approaches offer themselves for use here. One could use the "bridge" method with the measuring diode set up in a bridge circuit and the amplifier looking at the bridge signal.

However, a simpler approach is to use the amplifier as a differential amplifier which can compare two input voltages, one due to the diode and the other variable in order to select the point at which the scale of the instrument starts.

Such a circuit is shown in Fig. 1, where the diode D1 is connected to vary the inverting input of the operational amplifier whilst the non-inverting input is set by the potentiometer VR1. The choice of inputs is required to give an increasing output voltage for

COMPONENTS . . .

Resistors

R1 22kΩ R2 1·5kΩ R3 1·5kΩ

All 10% &W carbon

Potentiometers

 $\begin{array}{cc}
VR1 & 10k\Omega \\
VR2 & 22k\Omega
\end{array}$ skeleton presets

Diodes

D1 1N914 (or 1S914 or other silicon diode)
D2 BZY88 (5:6V) Zener diode

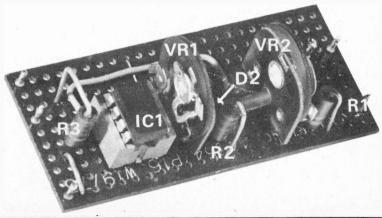
Integrated Circuit

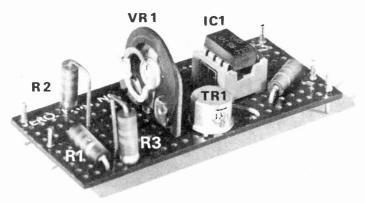
IC1 741 Operational Amplifier Miscellaneous

ME1 0-1mA Meter (see text) Veroboard (free with this issue) Total cost (excluding meter) about 95 pence.

Meters are available from about £2.50.







or its holder if one is used as in the prototype, and one to separate the output from VR1.

As can be seen, pins have been used to bring out the battery, input and output connections but of course wiring may be connected directly to the board if desired

Applications

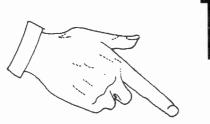
A circuit of this type can be used anywhere it is necessary to measure and control as a result of light variation. A side light control for a car is a typical idea. Other automotive applications include

light actuation of a garage door, car light failure alarm and perhaps even a burglar alarm.

Similarly in the home there are many applications starting off with things like a burglar alarm, light switches to turn house illumination on after dusk both for convenience and security and the like.

In technical areas devices of this sort can be used as object counters on a production line, as endstops for limiting machine travel, as automatic door openers and so on.

Dependent on the light resistor used the equipment will be sensitive to slightly different areas of the visible light spectrum but for normal visible light work most types will suffice including the well-known ORP 12 and its brothers from Mullard.



TOUCH SWITCH

Multi-purpose switching by touch alone

THE touch switch has many household, commercial and industrial applications. Specific examples in each area are intrusion alarms, opening supermarket doors and safety devices for machine operators.

In principle, simply touching a wire or metal plate will cause a relay to operate. To fulfil this the intermediary circuitry must be sufficiently sensitive to detect the very small noise currents provided by the hand.

Timer Chip

Touch switch circuits usually consist of cascaded amplifier stages (f.e.t. or bipolar). Construc-

tion of these for a beginner can be irksome and the results disappointing through spurious triggering, a common condition with these circuits

A convenient and economic package which can be simply adapted for this role is the Signetics 555 timer integrated circuit. Although this is usually used to provide accurate delay periods from microseconds to hours it will suit our purpose to use this facility to provide a variable period latch for the relay.

Input sensitivity is extremely high and the chip has a useful relay drive capability so that any relay coil operating in the 6-9V range and up to 200mA.

Simple Circuit

The first thing that one notices in looking at the circuit of Fig. 1 is its simplicity and economy in parts which makes for easy construction.

Since the internal functioning of the 555 was more than adequately covered in the June 1973 issue it will suffice to say that the circuit operates as a sensitive monostable.

In the quiescent state, that is with no hand applied to the touch plate or wire, battery drain is about 7mA none of this being taken by the relay. When a hand is applied relay coil current flows and the contacts close for a period determined by the timing constants R1 and C1.

For the component values shown the time is around 3 seconds but longer delay periods can be achieved by increasing the value of C1. With the resistor given delay periods in seconds for a particular capacitor can be approximated by the simple equation $t=9C_1$ where C_1 is in microfarads. For a delay of an hour you would, for example, need a $400\mu\text{F}$ capacitor but leakage would no doubt affect this estimate. A practical limit for very long delays would be about $10.000\mu\text{F}$.

Triggering

The current required at pin 2 to activate the circuit need only be half a microamp which indicates the extreme sensitivity. Unfortunately this can create problems of backlash triggering particularly with inductive loads. This is obviated by the inclusion of R2

The transient suppressing diode protecting the chip can in fact be any switching diode.

Touch Off

There are times when a touch on/touch off switching facility might be required as with say, a bedside night light. For this do

COMPONENTS . . .

Resistors

R1 8·2MΩ R2 2·2MΩ

All 10% ½ watt carbon

Capacitors

C1 0.33μ F polyester (see text)

Integrated Circuit

IC1 NE555V (Signetics)

Diode

D1 0A47

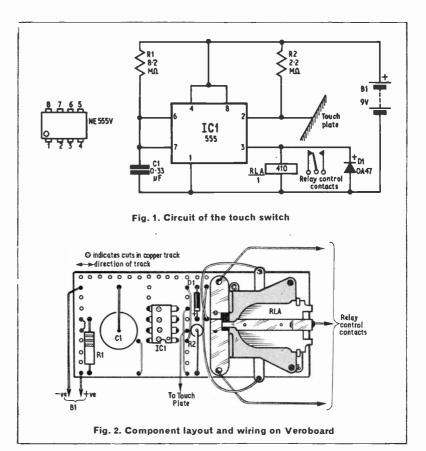
Relay

RLA 6V 410 Ω coil (type 912 R.S.)

Miscellaneous

Veroboard (free with this issue), 9V battery

£2 approx.



not connect pin 4 to the supply line but to another touch plate. A simple method of providing two adjacent touch plates is to join equal numbers of copper strips on a largish piece of Veroboard and simply wiring each "plate" to the appropriate 555 pin.

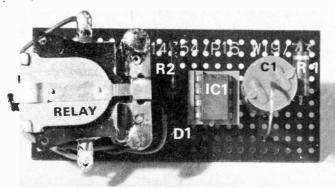
Construction

The components are easily mounted on the sample Veroboard as in Fig. 2. Cuts to be made in the copper underside are also clearly shown.

The 8 pin d.i.p. socket was included as it simplifies any future need to remove the chip either through damage or for use in another project

Other Relays

The delay periods provided by the timing component are almost independent of supply voltage. Since the 555 can be operated between 4.5 and 16V the supply rail can be lifted to accommodate other relay drive requirements but this should not exceed 16V and 200mA







RAPID, often startling, developments which have been made in the field of telecommunications owe much to the application of electronics. But electronics are, of course, contributing to the advance of other forms of human communication. Take, for example, one of the most traditional communications systems of all—the letter post. Electronics have an essential role in the multi-million pound plans to automate Britain's postal service.

THE NEED FOR AUTOMATION

The need to automate the postal service is imperative for two major reasons. First, to reduce the service's dependence on manpower (75 per cent of all Britain's postal costs are for labour) and secondly to maintain traditionally high standards while meeting the demands of today's fast moving world.

Since each of the 35 million letters the Post Office handles every working day in this country may have to be sorted up to eight times during its journey there is clearly ample scope for mechanisation and, helped by electronics, much has already been achieved. Electronics systems are enabling machines to select, stack and sort letters automatically at speeds far higher than even the most skilful postman-sorter.

The key to automation is the British postcode system, of its kind the most advanced in the world. Postcodes have been developed as the answer, or at least a supremely effective compromise, to the so far insuperable problem of total automation through machine-read addresses. Research is progressing but apart from the obvious difficulties with machine-reading of handwritten addresses it will still be many years before even typescript addresses can be read reliably without unreasonable restrictions on inks or type faces.

The postcode, however, by expressing an address in a machine language, easily keyed and translatable electronically, enables a high degree of automatic sorting to be introduced with present technology.

THE PRESENT SITUATION

Fully mechanised sorting offices are now operating at 12 centres in Britain and the eventual total will be around 100. Progress and the eventual success of the system depend on two factors—the completion of the programme to give each of Britain's 20 million addresses its own postcode; and the subsequent use of the postcodes by the public. In fact, postcoding of the country will be completed by August this year and while use of existing postcodes is rising steadily the importance of maximum use will become clear from the remainder of this article.

THE POSTCODE

The postcode is familiar to most people but the reasoning behind its planning and form is not so widely known and its explanation is appropriate here.

In fact the principle is simple. Each code consists of between five and seven alpha-numerical characters. The first one or two alphabetic characters in the code refer to one of 121 main towns or code centres in the UK (e.g. S for Sheffield, IP for Ipswich, LS for Leeds). A number following the alphabetic characters in the first or "outward" part of the code represents a sub-division of the area (e.g. IP3, IP7, IP17). These districts are geographic units similar in concept to the familiar London postal districts, such as SE1.

Each district is sub-divided into smaller areas or sectors which are again geographic units, represented by the first character or "inward" part of the postcode (e.g. IP3 9--, LS8 5--, etc.). Finally, the last two characters of the postcode give information about a street, part of a street, or, in the case of larger firms, a single address.

Thus a complete code in the case of part of a street in Ipswich for example reads IP3 7PS with the final two letters, PS, pinpointing that part of the street—usually between 15 and 20 addresses.

INITIAL SORTING

During initial sortation specially trained operators at postcoding desks equipped with electronic keyboards read each code and print onto the envelopes a series of code marks. The complete code is translated into a pattern of up to 28 marks and printed in two rows of 14. The code marks are virtually invisible to the naked eye but glow blue under ultraviolet exposure (see Fig. 1). The glow remains for a time after the ultra-violet light has been removed and this after-glow is detected by a photo-multiplier tube and the pattern recovered.

The pattern is next translated electronically to decide into which of 144 sorting boxes the letter is to be deposited by the machine. Once the postcode has been transcribed at the coding desk repeated exposure of the code marks and subsequent reading allow fully automatic sorting of the letter during its journey until it reaches the postman who is to take

it on his delivery round.

FOUR PROCESSES

While looking in closer detail at the electronics behind the coding desk and the automatic sorting machine it is necessary to follow in sequence the

^{*}Post Office Mechanisation Branch

four processes which a letter goes through from being received in the sorting office and being slotted automatically into its sorting box on the machine. The first two processes—segregation and facing—are preparatory to the final two—coding and sorting.

When letters arrive in a fully mechanised sorting office those which can be handled by automatic sorting machinery must be segregated from those which cannot. A wide assortment of letter shapes and sizes are found in every mail collection and those that are too unwieldy or thick must be culled from the letter stream before being allowed to pass into and possibly jam, or severely damage, the automatic sorting equipment.

The letters pass through a series of ingenious processes—rotating slatted drums, tilted conveyor belts, spaced rollers—which between them allow the unsuitably sized letters to slip from, or be plucked out of the stream, collected and then sorted by hand. The remaining letters are stacked automatically and passed to the next machine—the automatic letter facer (ALF). This is the first of the automatic sorting stages to employ electronics (see Fig. 2).

AUTOMATIC LETTER FACER (ALF)

The job of ALF is to arrange the letters so that they face the same way (the stamp upwards) preparatory to the stamps being postmarked or cancelled. Additionally ALF separates First and Second class letters so that the former may be sorted later.

While the operation of letter sorting machinery relies on the glow picked up from the code-marks printed onto the envelopes at the coding desk, ALF relies on a similar glow effect but obtained from the stamps themselves. Each stamp is surfaced with bars of phosphorescent material. The number of bars differ according to the value of the stamp. This enables the machine to detect if a stamp is present on the envelope, its position and whether the letter is being sent First or Second class.

ALF first strips letters off the feed stack and accelerates them along a series of rollers to obtain a stream of letters with a small gap between each one. The letters then pass under the first of two arrays of ultra-violet (UV) lamps which causes the phosphor on the stamps to glow. The glow continues after the letter has left the UV section and passes under two photo-multiplier tubes.

One of the tubes scans the top section of the letter face (which was exposed to the UV) and the other tube looks at the bottom of the face. In the dark only a tiny current flows in the multiplier but the glow from a UV "excited" stamp boosts the current.

The change in the current is detected and if a stamp is "seen" at either end of the letter face the letter is allowed to pass on. If no glow is picked up ALF assumes that the letter is facing the wrong way up and routes the letter through a twisted belt section to turn the letter over.

Now, all facing the same way, the letters pass under a second series of UV lamps and photomultiplier tubes. This time the number of phosphor bars on the stamps are noted and the letters identified as First or Second Class. At this stage, also, underpaid letters and those bearing no stamps are spotted.

ALF actually sorts letters into five boxes (cancelling their stamps on the way) of which two are for First Class, two for Second Class and the fifth for underpaid letters. Pairs of boxes are needed so



Fig. 1. A typical postcode pattern. The dots imprinted at the coding desk are normally almost invisible to the naked eye and have been outlined here

that letters with stamps in the leading position can be separated from those with stamps in the trailing position. The operator who unloads ALF correctly combines these two stacks before passing them to the next stage, the postcoding desk.

CODING DESK

The Coding Desk, with the help of an operator, provides each letter with its pattern of UV sensitive marks translated from the postcode. Where letters do not carry a postcode the operators imprint a code giving the post-town destination of the letter which enables initial automatic sorting to go ahead. But without the complete information which the correct postcode provides these letters must pass through a second coding desk to identify the delivery street

Fig. 2. Inside an automatic letter facing machine (ALF) which arranges letters so that they face the same way and the correct way up in the sorting office. ALF also separates first and second class mail and cancels the stamps, dealing with up to 20,000 items of mail an hour

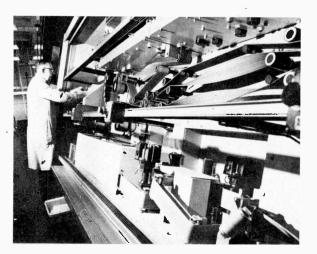


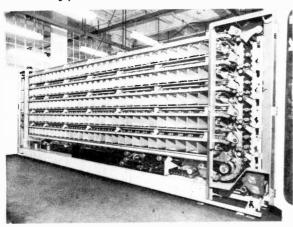


Fig. 3. Letters passing through a postcoding desk operated by a specially trained postman

when they arrive at the delivery sorting office. The full potential of the postcode is therefore missed and hence the importance of using the postcode wherever it is known.

The coding desk presents each letter at a display "window" facing the operator who is seated at a keyboard (see Fig. 3). The operator, who is a specially trained postman, types out the postcode on his keyboard which converts each character into a different pattern of binary signals on six wires.

Fig. 4. A medium speed Letter Sorting Machine with 144 selection boxes. These automatic machines read the first half of the postcode and sort the letters to their office of destination. At this office coded letters are again automatically sorted by machines which read the second half of the code and sort for the appropriate delivery postman



ELECTRONIC TRANSLATOR

The signals from each coding desk keyboard are sent to an electronic translator which contains a memory store of the patterns sent out from the keyboards. The translator recognises when the operator has completed a postcode and converts it into a pattern of 28 binary digits (bits). This 28 bit answer is then transmitted to the coding desk. Once this answer is received the coded letter is moved on to the print position.

Because the code marks are indelible the letter is held for a time in a cancel position so that the operator has time to correct any mistake before the

irrevocable printing step is taken.

The information received by the translator on the code pattern to be printed applies to the letter held in the "cancel" position and must be remembered until the letter is printed. Thus two translations must be stored in the desk and kept in synchronisation as the letters pass through. From being first presented in the display window to printing the average time the letter spends in the coding desk is two seconds.

After printing, the letter is ejected and collected and stacked automatically before being passed through the sorting machine. (In delivery offices post-coded letters arriving from other offices have already been through the coding desk stage at the previous office and are simply sorted automatically down to street or single address level. Here the last part, or "inward" code is used).

THE LETTER SORTING MACHINE

Like the coding desk and ALF the sorting machine (see Fig. 4) is fed letters in stacks and has first to separate them into a stream of spaced letters. They are then passed under UV lamps to generate the glow from the imprinted code-marks which are detected by the code-mark reader.

In the reader the row of marks on the letter pass a narrow slot behind which is a photo-multiplier tube which "recognises" each mark it sees by the increased current caused in the tube. The coded marks are read and stored in the sorting machine's memory. The code is fed into a second electronic translator which returns a 15-bit pattern identifying the box to which the letter should be routed.

ROTATION

The machine continues to store the translation and passes it on in stages for every two inches of letter travel. About three feet after passing the codereader the letter is rotated through 90 degrees to travel the remainder of the machine with its longest edge forwards.

Just before the right-angle rotation a photo-electric beam is broken to give an exact positional reference of the letter within the machine. The original codemark translation continues to be passed on its stages while the machine keeps track of the letter. When it arrives at the appropriate sorting box the box is opened and the letter has been sorted.

If the sorting machine is being used for outward mail (that is letters which are to go on to another sorting office) the letters are taken from the boxes, bundled and sent on by road or rail in the normal way. If the machine is sorting letters which are to be delivered in the sorting office area (inward mail) the letters are taken from the boxes and prepared for actual delivery by the postman on his round.

SYNCHRONISATION

All three of the machines described in this article, ALF, the coding desk and the sorting machine, must perform the required operation on each letter passing through the machine at the correct relative time.

To achieve this on the sorting machine and ALF it is necessary to pick up a signal to keep the electronics in synchronisation with the machine and hence the letters moving through it. This is achieved

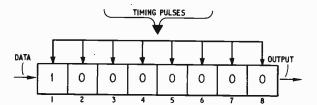


Fig. 5a. Diagrammatic representation of a shift register. The timing pulses cause the data held in each of the bistables to be shifted one place to the right

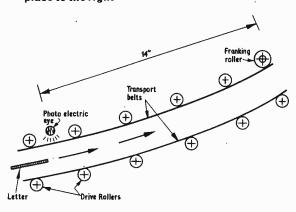


Fig. 5b. After eight shifts the letter will have reached the franking roller and a logic 1 in position 8 of the shift register causes the stamp to be cancelled.

by shining a light beam through a series of holes drilled around the perimeter of a wheel driven by the machine's transport belts. A phototransistor is illuminated by the beam whenever a hole passes. The current changes in the phototransistor are detected and used to generate timing pulses which step data through electronic models of the machines and hence control the letter flow.

All the control functions of the machines follow a simple scheme. Somewhere on the machine some characteristic of the letter is inspected and the signals generated are converted to logic voltages. From this information a decision is taken as to which of a number of operations the letter should undergo. However the alternative operations may be carried out some distance down the machine and it is necessary to keep track of the letter and its data until the point of operation is reached.

SHIFT REGISTER

This letter tracking is achieved by storing the data in a shift register and moving it one place down the machine for each timing pulse received. Thus on the sorting machine and ALF each stage of the shift register represents information associated with a letter whose leading edge is in a particular 2in length of belt.

At some later stage on the machine, where the operation has to be carried out, the associated shift register stage can be looked at electronically to detect when the letter is present and then indicate the operation. The word carried down the shift register varies in complexity from a single bit on part of ALF indicating whether or not a letter has to be turned over, to the 15 bits identifying the sorting machine outlet.

The Shift Register has a vital role to play in postal machines; repeatedly information is collected some distance from the position where action is taken on it.

In simple terms, let the figure represent an 8 stage shift register which represents the occupancy of 16 inches of a machine's transport belts. When no letters are present all the stages of the shift register are at logic "0" (Fig. 5a). A photo-electric eye detects the leading edge of a letter and marks the first stage of the shift register at logic "1" so it now reads 100 000 00. Each timing pulse from the machine passes the information on stage down the register. So, after 7 pulses, stage 8 alone will be marked 000 000 01 if no more letters have entered this section of belt.

A logic circuit is used to detect stage 8 being switched to logic "1" and causes the franking roller (14 inches from the photoelectric eye) to operate, cancelling the stamp (Fig. 5b).

Shift registers can be made to carry much more information and this need not all be collected at the same time.

ELECTRONIC CIRCUITS

From the "job descriptions" of the machines involved in automated mail handling it is clear that they rely heavily on electronics for both machine timing control and data storage.

All the electronic circuits required are constructed from components soldered onto small plug-in printed circuit boards. The components used—silicon planar epitaxial diodes and transistors, metal oxide resistors and polystyrene capacitors—are all highly reliable. However reliable, of course, some failures are inevitable, in which case a new circuit can readily be inserted in exchange.

Many thousands of these circuits are in use by the Post Office throughout the country. Although more than a hundred different circuits have been developed for various tasks, perhaps a dozen commonly-used circuits form some 90 per cent of the total.

DIGITAL LOGIC

Most of these common circuits are digital logic circuits requiring inputs of either +6.6 volts (logic "0") or 0 volts (logic "1") and giving outputs at the same levels. Some of the circuits, however, have to interface between the logic system and the electromagnetic devices used to control the machine. One

of these gives a logic output when the current in a phototransistor exceeds a certain level; another has to apply voltage to a load device (which could be a diverter or print pin solenoid) whenever its input is at 0 volts. However, not all circuits operate with logic levels at either input or output.

DOLLILOG

The range of logic circuits is often referred to as "Dollilog". This name derives from "Dollis Hill Logic" after the PO Research Station at Dollis Hill where the system was developed.

Some of the more common Dollilog circuits are described briefly in the following paragraphs together with some ideas on how they are used.

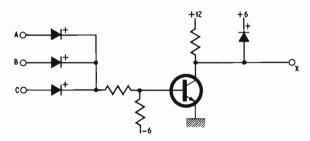


Fig. 6. Circuit diagram of a three input NAND gate used in the Dollilog system

NAND GATE

The diagram (Fig. 6) shows a three input NAND gate. The output (x) of this circuit will be at the logic "0" voltage (+6.6 volts) only if all three of the inputs (A, B and C) are at the logic "1" voltage (0 volts). In this case the transistor will be cut off.

volts). In this case the transistor will be cut off.
If any one of the inputs is at the logic "0" voltage
the transistor will be saturated and the output will
be at the logic "1" voltage.

The NAND gate can be used to detect when several different conditions occur simultaneously and cause some process to be carried out.

Other logic circuits are used, one of which only gives the logic "1" output voltage when all three inputs are at logic "0", this being a NOR gate. Another circuit gives a logic "1" if both inputs are the same.

SCHMITT TRIGGER

The Schmitt trigger has two outputs which are never the same. The outputs states interchange when ever an input voltage exceeds one threshold or falls below another.

POWER AMPLIFIER

The power amplifier's output transistor will be saturated whenever the input to the circuit is at logic "I". The usual circuit can control loads of up to 25 watts although other high power versions exist.

BISTABLE MULTIVIBRATOR

The bistable multivibrator is another circuit which has two outputs which are always opposite. The outputs are switched by pulses applied to "set" and "reset" inputs. Bistables can be used to store information.

SHIFT REGISTER

Shift registers are arrays of bistables with a common clock input which can be pulsed to step data along the register.

NEW TECHNOLOGY

To some readers, these circuits will look a little dated. To some extent this is true; it must be remembered, however, that the system has proved satisfactory in use. The Post Office has to consider any change on this scale very seriously. Various factors have to be considered such as replacement cost, the cost of training maintenance engineers in any new system and the compatibility of any new equipment with the old.

At present, many new techniques are being tried and tested. The coding desk keyboard contains a number of DTL integrated circuits while the new memory which will replace the sorting machine's original mechanical memory is being developed using MOS integrated circuits. A new translator uses small general purpose computers rather than specially built electronic circuits. At present, however, only the relatively cheap keyboard, where size is critical, uses integrated circuits in a production item.

*

EE POINTS THE WAY

IN THE NOVEMBER ISSUE — AN 8 PAGE SUPPLEMENT

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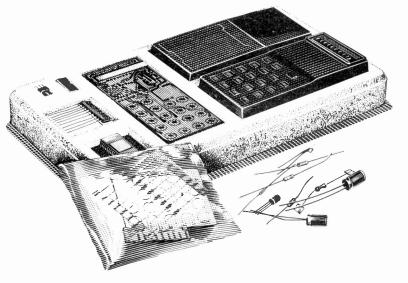
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- 2. Large-scale integrated circuit.
- 3. Interface chip.
- 4. Thick-film resistor pack.
- Case mouldings, with buttons, window and light-up display in position.
- 6. Printed circuit board.
- 7. Keyboard panel.
- 8, Electronic components pack (diodes, resistors, capacitors, transistor).
- Battery clips and on/off switch.
- 10. Soft wallet.



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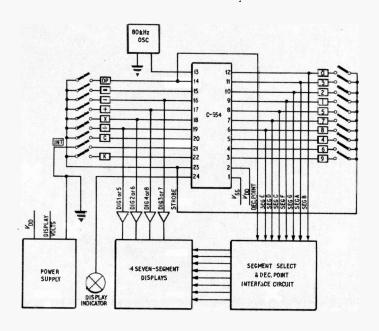


Fig. 1. Block diagram of a four digit calculator using the C554 integrated circuit

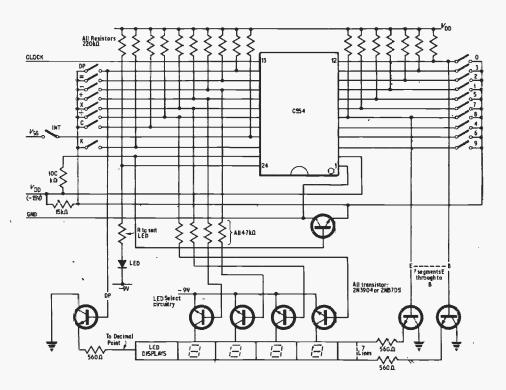


Fig. 2. A more complete circuit diagram showing l.e.d. seven segment display with all the necessary driver transistors. All voltages are with respect to the V_{SS} line

DEVICES ... APPLICATIONS

FOUR DIGIT MOS CALCULATOR I.C.

T SEEMS only a short while ago that we were all impressed by the technology needed to put a complete calculator on one chip, yet already General Instrument Microelectronics have developed a second generation M.O.S. calculator integrated circuit. The main advantage of the new device is greatly reduced power consumption, this being made possible by reducing the necessary supply voltage from 25 volts to 15 volts.

The new calculator i.c.s, type C550 and C554, enable either eight (C550) or four (C554) digit calculators to be built. The reduction in power supply voltage is made possible by the use of new materials. The new process is called General Instrument Advance Nitride Technology (GIANT). Reduction of power supply voltage reduces chip consumption to a mere 150mW.

FOUR DIGIT CALCULATOR

Since eight digit calculator i.c.s are most widely used the more unusual four digit i.c. (C554) will be described. This calculator needs only a four digit display so is much cheaper and probably just as useful as the eight digit type.

Fig. 1 shows the block diagram of a calculator built around the C554. All switches would be contained on a keyboard. Some of the connections to the i.c. are used both as inputs and outputs, this being done to keep the pin count down to 24. The i.c. is driven by an 80kHz clock generator.

The i.c. generates a strobe pulse at pin 23 which is fed to all the keyboard contacts except the interchange (INT) switch. If any of the keyboard contacts are closed during the strobe pulse then the appropriate input will be fed to the i.c. The strobe pulse only occupies a very small proportion of the total clock period, the rest being used to display the output.

DISPLAY

A multiplexing system is used for the output whereby each of the four digits is selected in turn. A seven segment output is generated on the same lines used as numerical inputs (pins 6 to 12) as well as decimal point information. The display can be any seven segment type such as l.e.d., liquid crystal or hot filament; the driver (interface) circuit is the only thing that needs changing.

Fig. 2 shows a more detailed circuit diagram using l.e.d. seven segment displays and single transistor drivers. The $220 \mathrm{k}\Omega$ resistors are needed because the MOS output transistors are "open drain" types. All voltages are specified with respect to the $V_{\rm ss}$ level. The 560Ω resistors are simply to limit the current through the l.e.d.s and the $4.7 \mathrm{k}\Omega$ resistors are to limit the transistor base current.

INTERCHANGE SWITCH

One point that has not been so far explained is the interchange switch labelled INT. The C554 has been described as a four digit device in that it only displays

four digits at a time but, in fact, it can accept up to eight digit inputs and produces an eight digit output. The interchange switch interchanges the two sets of four digits so that more accurate answers may be obtained if desired. The l.e.d. shown indicates which of the groups is being displayed. The decimal point appears in whichever group is appropriate: if the decimal point does not occur in the group being displayed it will be obvious as all the decimal points on all four digits will be illuminated.

As mentioned earlier the C554 needs a 15V power supply and an 80kHz clock input. Both of these needs can be satisfied with the circuit shown in Fig. 4. An oscillator generates the 80kHz clock pulses and its output is fed to a regulator which gives 15V at up to 10mA.

USING THE CALCULATOR

The C554 can perform all the normal arithmetic operation: addition, subtraction, multiplication and division. The true sign of answers is given if the result is negative the minus sign replacing the most significant digit so that only three digits are available.

There is a single CLEAR key which can be used to clear either the complete calculator or the last entry. If the depression of the CLEAR key is followed by pressing a function key, only the last entry will be cleared; if a digit key is pressed then the whole calculator will be cleared.

There is a CONSTANT facility whereby a result of a calculation might be stored to be used after the entry of another number. The constant store can hold a multiplier, divisor, subtrahend or addend.

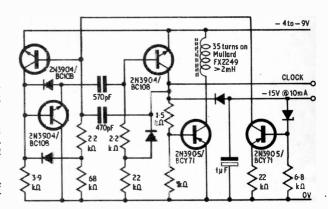


Fig. 3. A power supply and clock generator for the C554 i.c. which is all that is needed to complete the calculator in Fig. 2



COLLOWING the circuit description last month the construction of the semi-conductor Tester will be described.

CONSTRUCTION

Most of the components are mounted on 0.15in matrix Veroboard measuring $3\frac{5}{8}$ in \times $1\frac{1}{4}$ in. Breaks in the copper strips should be made as shown in the Veroboard layout of Fig. 7.

The positions of the components are shown the reverse of Fig. 7, but the layout is not critical. The resistors and capacitors are mounted first, remembering to connect the electrolytic capacitor C6 $(2\mu F)$ the correct way round i.e. +ve to +ve. The component leads are fed through the Veroboard holes and bent so as to lie along the copper strips. The leads are then cut so that about 0-lin of lead remains on the copper strip, and then the leads are carefully soldered to the strip.

The transistors and diodes are mounted last, care being taken not to overheat the leads as this can cause component failure. Meter face

The meter face should be modified to include all the numbers in Fig. 8. Letraset can be used for this purpose.

Case and component interconnections

The case used was a 18cm \times 15cm \times 13·5cm West Hyde MOD303.

The lettering on the front panel may be done using Letraset. The components are then mounted on the front panel and the connections shown in the circuit diagram are made. Use a different coloured wire for each connection link if possible for simplicity.

Connect one male battery connector to a female battery connector by 4in of wire for the battery "jumper" lead.

VEROBOARD PANEL

Printed Circuit Supporting Bracket

The bracket used to support the Veroboard panel may be made from aluminium sheet, 22 s.w.g. being quite suitable.

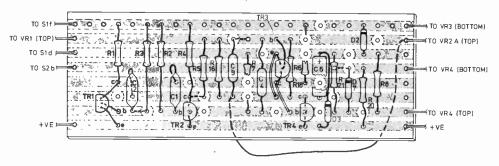


Fig. 7. Layout of the components on the Veroboard panel

Two holes should be made in the Veroboard panel

to align with holes made in the bracket.

The batteries are retained in the case in the prototype using one piece of 1in wide elastic which is 8.5in long. Each end of the elastic is secured to the base of the case by a 4BA bolt, and nuts. The feet supplied with the case are now stuck to the base of the case. The test leads are made as shown in Fig 9, each lead being about 60cms long.

TESTING

Testing may be accomplished as follows, the components being tested are listed at the beginning of each test to facilitate fault finding by beginners.

- S3, Batteries, S1A, S1B, S1C, R9, D4, D5, 52A, 52B, 526, ME1, R14, S4. Socket leads C and E.
 - (a) Set METER switch to I_{ceo} , DEVICE switch to PNP, and ON/OFF switch to ON:—meter should indicate zero.
 - (b) Connect a resistor of approximately 50 k Ω between leads C and E:—meter should read approximately $200\mu A$. (f.s.d. = 10mA).
 - (c) Depress F.S.D. ÷ 20 switch:—meter should read approximately 200µA.
 - (f.s.d. = 0.5mA). Release F.S.D. ÷ 20 switch. (d) Set DEVICE switch to NPN and repeat 1(b)
 - and I(c).

 (e) Connect lead C to lead E:—meter should read
 - (e) Connect lead C to lead E:—meter should read approx. 10mA (f.s.d. = 10mA).
 - (f) Set (METER switch to I_c and repeat (e).
 - (g) Disconnect lead C from lead E:—meter should read zero.
 - Switch OFF tester.
- R1 to 6, R16, R21, R18, R17, VR2A, C1, C2, C3, C6, D3, D7, TR1 to TR4, S1F, S2A, 52C, Meter, R7, C4.

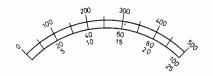


Fig. 8. The modified meter face

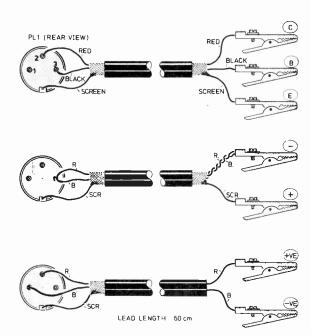
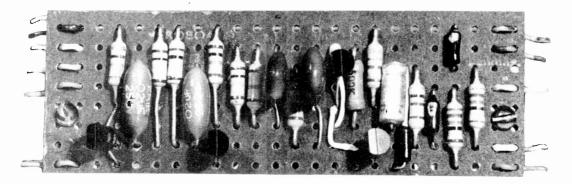


Fig. 9. Transistor, CSR and diode leads

Rear view of the front panel of the Semiconductor Tester



Photograph of the components on the Veroboard panel

- (a) set METER switch to h_{fe}, DEVICE SWITCH TO PNP, ON/OFF switch to ON:—meter should read approx. f.s.d. If meter reading is somewhat less than f.s.d. decrease value of R16, if meter reading is greater than f.s.d. increase value of R16. (b) Set DEVICE switch to NPN:—meter reading
- should remain as in 2 (a). Switch tester OFF.

 3. VR1, R15, R19, C3, R22, S1E, S2D, VR2B, R10. (a) Connect a "working" transistor to the transistor socket SK1 or the "Flying Leads" lead C to collector, lead B to base, and lead E to emitter. Set METER switch to I_c, DEVICE switch to PNP or NPN depending on transistor type, set SET I_c to min, switch ON tester:—meter reading should increase as SET I_c control is turned from min. to max. Adjust SET I_c so that the meter current is 2mA (f.s.d. = 10mA).
 - (b) Set METER switch to h_{fe} and adjust h_{fe} control:—it should be possible to find a position where the meter deflection is a minimum. Switch OFF tester and remove the transistor.

4. S1F, LP1, D2, R8, VR4, R20, VR3, D1, R11, S2A, R13, R12, S2C, S4, S2B, S1E, R22, S2D, D6,

R23, leads b.

(a) Set DEVICE switch to THY ZEN UJT V. METER switch to THY GATE 1, turn SET I to max, and SET V to min. short leads B and C, switch ON tester and adjust SET V from min. to max.: -meter reading should increase from zero to approx. $\frac{3}{4}$ f.s.d. (b) Set SET V to min, depress and hold F.S.D. \div

20 switch and adjust SET V for f.s.d., release F.S.D. ÷ 20 switch:—meter should now be 1/20 of f.s.d. Remove short between leads B and C,

switch OFF tester.

(c) Set METER switch to THY GATE V, switch ON tester, adjust SET V from min. to max:meter reading should decrease from 12V to 0V (f.s.d. is 25V)

(d) Set METER switch to THY HOLD I, turn SET I to min, short leads C and E, and increase SET I from min. to max: - meter reading should increase to approximately f.s.d. when lamp L1 should be lit.

(e) Set SET I to min., depress and hold F.S.D. ÷ 20, increase SET I control until meter reads f.s.d., release F.S.D. ÷ 20:—meter should read 1/20 f.s.d.

OPERATING INSTRUCTIONS

Transistor testing

- 1. Connect transistor to socket SK1 or to flying leads, E to emitter B to base, C to collector.
- Set DEVICE to PNP or NPN depending on transistor type).
- 3. Set METER switch to I_{ceo} , switch tester ON—meter shows I_{ceo} (f.s.d. 10mA). Operate F.S.D. \div 20 if required.

4. Set METER switch to I_c .

5. Adjust SET I_c control for required I_c on meter (f.s.d. 10mA).

6. Set METER switch to h_{fe} .

7. Adjust h_{fe} control for minimum reading. The h_{fe} is indicated on $h_{\rm fe}$ scale.

Diode Testing

1. Connect to socket SK1 on to flying leads of transistor lead. E to anode and C to cathode.

- 2. Set METER switch to I_{ceo} . 3. Set DEVICE switch to PNP, switch TESTER ON meter shows forward current (f.s.d. 10mA) RF = 12000 — 1**20**0. Ir (in mA)
- 4. Set DEVICE switch to NPN—meter shows reverse current

12000 -1200 (F.S.D. $\div 20$ may be $\overline{I_R}$ (in mA) used if required

Zener Diodes (0-10V)

- 1. Connect diode to socket SK1 or to flying leads of transistor lead. C to cathode E to anode, connect lead B to lead C.
- Set DEVICE switch to THY ZEN UJT V.
- 3. Set SET V to min.
- 4. Set SET I to min.
- 5. Switch METER switch to THY GATE V/V. switch tester ON—meter reads Zener voltage (f.s.d. = 25V).
- 6. Adjust SET I from min. to max. meter reading should only change by a small amount. This check gives a rough indication of the a.c. resistance of the diode. (The current through the diode

at any point can be found by setting meter switch to THY HOLD I, I.

Note that for Zener voltages in excess of 10V it is necessary to insert a battery of known voltage between the emitter lead of the Zener diode and lead E of the tester, then Zener Voltage = meter reading + battery voltage.

Thyristor Testing

- 1. Connect the thyristor to SK1 or to the flying leads of the transistor testing lead anode to E, cathode to C, Gate to B.
- 2. Set SET V to min.
- 3. Set SET I to max.
- 4. Set DEVICE switch to THY ZEN. UJT. V.
- Set METER switch to THY GATE I, switch tester ON.
- Adjust SET V slowly clockwise noting the meter current. When the lamp L1 lights (thyristor conducting). Use FSD ÷ 20 if required. Meter reading to gate trigger current.
- 7. Set SET V to min.
- 8. Switch tester OFF then ON—lamp L1 should extinguish indicating that thyristor has reset.
- Set METER switch to THY GATE V—meter indicates supply voltage.
- 10. Adjust SET V slowly clockwise—meter reading slowly falls then suddenly drops to zero, note the voltage (trigger voltage) at which the meter reading suddenly falls.

Gate trigger voltage = Supply voltage - trigger voltage

- 11. Set SET V to min.
- 12. Set METER switch to THY HOLD I, I.
- 13. Adjust SET I from max. to min. and note the point at which the meter current falls to zero, this current is the thyristor holding current. (F.S.D. ÷ 20 may be used if required).

Unijunction transistor testing

- 1. Set DEVICE switch to THY, ZEN, UJT, V.
- 2. Set METER switch to THY GATE V, V.
- 3. Set SET V to min.
- 4. Connect unijunction to SK1 on to flying leads Base 1 to C, Base 2 to E, Emitter to B.
- 5. Switch tester ON.
- 6. Adjust SET V (noting decrease in meter voltage reading until meter "kicks" when:—

Intrinsic Standoff ratio = $\frac{\text{decrease in meter voltage}}{\text{maximum meter voltage}}$

Using tester as a Voltmeter

- 1. Set DEVICE switch to PNP.
- 2. Set METER switch to THY GATE U, V.
- 3. Connect voltmeter lead to voltage to be measured

Using tester as an Ammeter (f.s.d. = 100 mA) or 5 mA).

- 1. Set METER switch to THY HOLD I, I.
- 2. Connected current lead to current source (F.S.D.
 ÷ 20 may be used if required).

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	(Watts)	lb	oz				£	Þ
07	20	- 1	8	7.0 ×	6.0 ×	6.0	1.94	30
149	60	3	12	9.9 ×	7.7 ×	8.6	2.88	36
150	100	5	8	9.9 ×	8.9 ×	8.6	3-16	52
151	200	8	0	12·1 ×	9.3 ×	10.2	5.31	52
152	250	13	12	12.1 ×	11.8 ×	10.2	7.01	67
153	350	15	0	14.0 ×	10.8 ×	11.8	9.40	82
154	500	19	8	14.0 ×	13.4 ×	11.8	13.55	*
155	750	29	0	17·2 ×	14.0 ×	14.0	19.26	*
156	1000	38	0	17·2 ×	16.6 ×	14.0	24-97	*
158	2000	60	ŏ	21-6 ×			41-25	*
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				Α	UTO	SE	RIES		200	
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No.	(Watt	s) Ib	OZ						. £	P
113	20	1	0	5.8 ×	5.1 ×		0-115-21		1.02	27
64	75	2	4	7.0 ×	6.7 ×		0-115-21		2.00	30
4	150	. 3	4	8.9 ×	7.7 ×	7.7	0-115-20	0-220-240	2.42	36
66	300	6	4	9.9 x	9.6 ×		7.		4.70	52
67	500	12	8	12-1 ×	11.2 ×	10.2			6.98	67
84	1000	19	8	14.0 ×	13.4 ×	14.3			12.69	82
93	1500	30	4	14.0 ×	15.9 ×	14.3		11	18-39	*
95	2000	32	0	17.2 ×	16.6 ×	14.0			24-00	*
73	3000	40	ō	21.6 ×	13-4 ×	18-1			32-67	*

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107	6.0	12 0	14.0 ×	10.2 ×	11.8	.,		8.63	6
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124	0.5) 4	7·0 ×	6.7 ×	6-1	U-49-3U-4	U-70-06 V	1.07	

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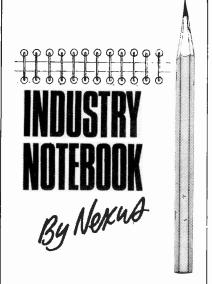
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TOUGH FIELD

It takes a lot of guts to take on the big boys, especially in a field as tough as mobile radiotelephones. This is the field where household names like Cossor, Ultra and STC have all chucked in the sponge within the past year or so.

Pye is still undisputed market leader and Burndept, strengthened by the acquisition of Cossor and Ultra business, is second. The question is whether Dymar can succeed when better-known com-

panies have failed?

I can report that there was no lack of confidence in the Dymar management when the company launched its new 'Lynx' range of mobiles on September 12. For one thing the Lynx, with a 20W output. has a technical edge on the competition and Dymar has prepared the sales ground well by appointing a nation-wide network of distributors and service centres.

Although Dymar as a name is not yet well known, the company has a huge back-log of experience in both design and manufacture of mobile radio. To date, Dymar has produced 30,000 mobile equipments of which half have been made under contract to other companies' designs and about 7,500 have been land or marine mobiles of Dymar's own design and manufacture.

Approximately £150,000 has been invested in the Lynx development and it could be a real winner. In fact the company expects no less than half the total 1974 turnover in Lynx with the other half split 35 per cent in marine radiotelephones, and 15 per cent in instruments of which Dymar make a range for manufacture, testing and servicing of r.f. systems.

SATELLITE NAVIGATION

While Dymar was launching Lynx in Whitehall, Redifon was launching marine satellite navigation receiver half-a-mile away in the Strand

The Redifon Satellite Navigator is another milestone along the road forecast by J. R. Brinkley when, earlier this year, he said the company was now fully committed to a huge modern technology programme to win a large share of the communications and navigation markets. It was on that occasion that he was able to announce that the Redifon Omega Navigator had already been fitted to 100 ships and had realised £0.5 million in revenue.

What industry observers are waiting for now is the Redifon v.h.f./u.h.f. programme. No hardware has yet been announced although R and D expenditure in this area is said to be £100,000. As part of the expansion programme a new plant is under construction in Wales which could ultimately house 1,000 workers.

BATTLE CONTINUES

The light-emitting-diode battle continues with increasing intensity. Ferranti's new green LED's are reported to be coming out of the factory at 1.5 million a year now, stepping up to a rate of 10 million a year by the Spring of '74.

Plessey's Optoelectronics Microwave Unit at Towcester is also in on the action with a new yellow gallium phosphide device which is claimed to have a higher light output and spot brightness than any other solid state light source currently available. The Towcester Unit is also offering custom-built hybrid arrays in green. yellow and red, a typical array recently produced for a customer 32 diodes. They are assembled and encapsulated on a metallised ceramic substrate which can include dropping resistors if required.

INSTRUMENTS— A HARD LIFE

instrument The industry struggling back to life after three years in the doldrums. Renewed activity springs from the big investment programmes now being undertaken by many major user companies.

But the instrument business is no sinecure. John Ceresa, marketing director of Racal Instruments, recalls that 13 years ago a 10MHz counter cost £400. By 1970, in ten years, the price had dropped 75 per cent for the same instrument in a solid state version and with much improved performance and reliability. You then found you had to

sell four times as many to make the same turnover and maybe eight times as many to make the same profit.

Another angle on the life and hard times of instrument men comes from David George, managing director of SE Laboratories, one of the companies who have been through a bad patch. He says it is suicidal for any company to try to compete in every generation of instruments. One of the ways of cutting R and D costs is to do a deal with a complementary manufacturer elsewhere in the world. SE Labs has actually done this with Data Precision Corporation in USA.

I note with interest that even Tektronix with a near stranglehold on the top end of the oscilloscope market is finding it hard to expand any further without diversifying other fields. Although the oscilloscope market shows no signs of a decline-it is still growing at least 10 per cent a year-Tektronix has moved into programmable calculators and more recently has acquired a Californian Company which specialises in TV studio equipment.

AD-POWER WORKS!

Just when you think there are no surprises left in the industry you encounter another. I had this experience when I visited Kingshill Electronic Products who make professional quality power supplies for the universities, the Government, and the industry in general. The company is a subsidiary of Islington Metal and Plating Works Ltd., one of the pioneers of chromium plating nearly 50 years ago. The electronics off-shoot was started in 1964 by Dr. Henry Herbst who still sits in the managerial chair. His sort of company would normally have a flock of travelling salesmen chatting up potential customers and a sales programme involving appearances at major electronic exhibitions.

Not Kingshill. Herbst has never had a salesman on the road, has never shown at an exhibition. Kingshill sells on press advertising and personal recommendation. It seems to work. The production lines turn out 18,000 units a year which not only go to U.K. users but also to out-of-the-way places like Fiji and Thailand.

The premises were a surprise, too. They were originally stables for the horses of the old London horse-drawn trams. Not so un-expected is that Herbst, with a solid business now established in power supplies and a firm economic base from which to operate, is looking for new electronic fields to conquer.

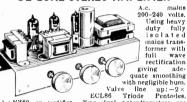
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BY J.B. DANCE

N LAST MONTH'S article the basic principles of phase locked loops were described. This month we will look at the application of these principles in f.m. radio reception, a.m. demodulation and trequency synthesisers.

RECEIVER TECHNIQUES

An amplitude modulated (a.m.) radio signal consists of a high frequency sine wave, whose frequency is fixed for any particular broadcasting station, but whose amplitude varies in sympathy with the audio signal that is the information to be transmitted. The carrier frequency f_s ranges from about 200kHz up to many megahertz, whilst the audio frequency ranges from 30Hz to 15kHz. The receiver must be able to reproduce the modulating audio signal as faithfully as possible and must provide as much separation as possible between signals whose carrier frequencies are close.

T.R.F. RECEIVERS

One of the simplest forms of radio receiver is the so-called "tuned radio frequency" (t.r.f.) receiver. The basic form of this type of receiver is shown in the block diagram of Fig. 2.1.

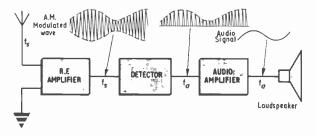
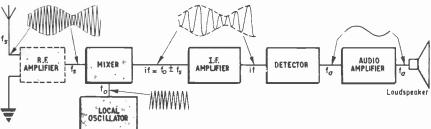


Fig. 2.1. A block diagram showing the stages in a t.r.f. receiver

Fig. 2.2. A block diagram of a receiver. The use of an r.f. stage is



The incoming signal is amplified in the radio frequency stage. No change in the frequency or in the form of the signal takes place in this stage, but tuned circuits are employed to provide selectivity and to reject unwanted signals of other frequencies.

The signal is then demodulated or detected, usually by a stage that provides some amplification. The audio signal from the detector (f_a) is fed to an audio amplifier which can provide sufficient power output to drive the loudspeaker.

The selectivity provided by receivers of this type is very limited, partly because it is not practical to employ more than about four tuned circuits operating at the radio frequency f_s . Each tuned circuit requires a separate gang of the tuning capacitor and the relatively long connections to this capacitor can lead to unwanted feedback and oscillation.

It is normally necessary to employ a fairly large value of tuning capacitor (typically 300pF to 500pF per gang) in order that the desired tuning range can be covered. This may make it impossible to design the tuned circuits for optimum Q factor (especially at high signal frequencies); the selectivity and gain therefore suffer.

The limitations of t.r.f. receivers become very noticeable at the higher frequencies. The pass band is so broad that it is impossible to obtain adequate selectivity and noise rejection, whilst the signal frequency amplification is normally inadequate in the short wave bands.

SUPERHETERODYNE RECEIVERS

Many of the disadvantages of the t.r.f. receiver can be avoided by the use of the superheterodyne principle, but some other undesirable features are then introduced. A block diagram of a superheterodyne receiver is shown in Fig. 2.2.

superheterodyne

ontional

The incoming signal from the aerial at a frequency f_s may be fed into a radio frequency stage (shown dotted) or directly into the mixer stage. In the latter stage it is mixed with a frequency f_o generated by a local oscillator. The frequency f_o is varied with the frequency to which the receiver is tuned.

The output from the mixer consists mainly of the sum and difference frequencies of the signals which are fed into it, that is, $f_o + f_s$ and $f_o - f_s$. It is arranged that the frequency of one of these outputs (normally the difference frequency in receivers covering the long, medium and short wave bands) is kept constant no matter what the frequency of the input signal, f_s . This constant frequency output signal is known as the intermediate frequency or i.f.

The mixer stage is followed by one or more stages of amplification at the intermediate frequency. These stages involve the use of a number of tuned circuits (typically six) which resonate at the intermediate frequency.

The intermediate frequency tuned circuits are fixed tuned and they can therefore be designed with the optimum inductance/capacitance ratio for high selectivity and gain; they provide virtually all of the selectivity of the receiver. Each tuned circuit can be incorporated inside a metal can so that unwanted coupling is reduced to a minimum.

DISADVANTAGES OF SUPERHETS

Superheterodyne receivers have the disadvantage that they tend to suffer from spurious responses. Noises such as whistles may be generated when the local oscillator frequency (or one of its harmonics) beats with certain incoming frequencies. One of these undesired responses occurs at the image frequency, where a signal at a frequency of $f_s + 2if$, mixes with the local oscillator frequency to form an interfering signal.

Image frequency interference may be reduced by using tuned circuits before the mixer or by using a high intermediate frequency so that the image frequency is far away from f_s and easily rejected by the tuned circuits in the r.f. amplifier, though this latter method tends to reduce selectivity.

Another disadvantage of superhets is that the tuned circuits used throughout cannot be formed on an integrated circuit chip. Integrated circuits are not only smaller but because of the ease of assembly they greatly reduce labour costs.

THE SYNCHRODYNE

The synchrodyne receiver is the forerunner of the phase locked loop and aroused great interest in the late 1940's. The synchrodyne type of receiver is illustrated in the block diagram of Fig. 2.3.

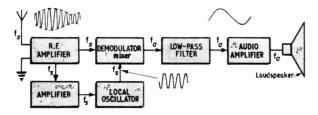


Fig. 2.3. The stages employed in a synchrodyne receiver

The signal from the aerial is fed via an optional radio frequency stage to the demodulator and also, via another amplifier stage, to a local oscillator. This signal keeps the local oscillator locked or synchronised to the input signal frequency—hence the name synchrodyne.

When the two signals of the same frequency come together in the demodulator stage (which is like the mixer of a superheterodyne), sum and difference frequencies are formed. The sum frequency, $2f_s$, is not required and is rejected by the succeeding low-pass filter. One might expect the difference frequency to be zero, but this really means that the carrier frequency is zero. In other words the output of the demodulator in Fig. 2.3 consists of the required audio signal minus the carrier.

Any interfering radio signals will not have the same frequency as the local oscillator. They therefore produce an output signal from the demodulator stage which is normally of a much higher frequency than that due to the wanted signal. The interfering signal cannot pass through the low-pass filter which follows the demodulator.

SYNCHRODYNE SELECTIVITY

One of the great advantages of the synchrodyne is that it does not obtain its selectivity by means of conventional tuned circuits, but by the use of a low-pass audio filter. This low-pass filter can be very simple (see Part 1) and can be designed to cut off at any point so as to provide the desired selectivity. It is easy to make the selectivity variable by adjusting the value of a single component.

Another important advantage of the synchrodyne receiver is that relatively few tuned circuits are employed in it. The only tuned circuit which is essential is the one in the oscillator circuit, but it is normally desirable to include some tuned circuits in the radio frequency stages to prevent very strong unwanted signals from overloading the mixer stage.

The local oscillator of a synchrodyne receiver must be locked in frequency to the wanted signal if one is to avoid the generation of beat frequencies. This type of receiver is either correctly tuned or not tuned at all. Unlike most other forms of receiver, a distorted audio output signal cannot be obtained as a result of the mis-tuning of the local oscillator.

As the local oscillator tuning is altered, the oscillator will remain synchronised to the incoming signal over a certain range, although the local oscillator signal may change in amplitude somewhat. Thus the only effect of a slight oscillator mis-tuning will be an alteration in the audio volume.

ADVANTAGES AND DISADVANTAGES

The synchrodyne receiver does not suffer from image responses. We have seen that an interfering signal at a frequency of twice the intermediate frequency plus the wanted signal frequency can produce an image response in a superheterodyne receiver; in the synchrodyne, however, the "intermediate frequency" is zero and therefore the image frequency becomes equal to the wanted signal frequency.

For domestic listening, the main objection to the use of the synchrodyne receiver is the loud whistles which are generated when one is tuning into a signal. These whistles are, of course, due to the local

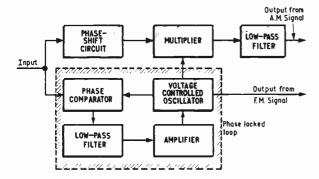


Fig. 2.4. An a.m. demodulator using a phase locked loop. The components of the loop itself are shown inside the box.

oscillator beating with the input signal before their frequencies become locked together. This problem can be overcome by the use of switched tuning.

Another reason for the limited use of synchrodyne receivers in the past has been the lack of simple, but effective phase locking circuits.

Other names for the synchrodyne receiver include "homodyne", "phase conrol" or "PC" receivers, synchronous detectors and phase locked detectors.

A.M. DEMODULATION

The block diagram (Fig. 2.4) shows a practical circuit for a.m. receivers using a phase locked loop.

The phase locked loop (comprising the units shown inside the dotted line) locks onto the input signal in the normal way. The voltage controlled oscillator produces a signal of the same frequency as the input signal, but it is unmodulated.

When the free-running frequency of the voltage controlled oscillator is the same as that of the incoming signal, locking will occur with these two signals 90° out of phase. In the system of Fig. 2.4, the phase shift circuit changes the phase of the input signal by 90° so that it is in phase with the output from the voltage controlled oscillator.

The two signals of the same phase are mixed in a second phase comparator or multiplier. The output from the multiplier is passed through a second low-pass filter so that only the difference frequency remains. This difference frequency is proportional to the amplitude of the input signal and is therefore the required audio output.

Systems such as that shown in Fig. 2.4 provide a higher degree of noise rejection than can be obtained from conventional a.m. demodulators which detect the peaked noise pulses.

F.M. DEMODULATION

The voltage controlled oscillator is normally designed so that its frequency is linearly related to the applied control voltage. As this latter voltage keeps the voltage controlled oscillator frequency the same as the incoming signal frequency, the control voltage is linearly related to the frequency deviation of the input signal.

In other words, the output voltage from the system (Fig. 1.1.) is the demodulated audio signal when the incoming signal is a frequency modulated one. The voltage controlled oscillator produces a greatly amplified copy of the input signal from which most of the noise has been removed.

One might expect that a very simple f.m. receiver could be constructed by merely feeding the input signal to the phase locked loop. In actual practice, however, a somewhat more complex system than this must be employed.

Currently available phase locked loops cannot operate at frequencies above about 60MHz. Most f.m. signals will therefore have to be converted to a lower frequency before they are passed to the phase locked loop. However, it seems likely that phase locked loops will be available at some future date which can operate at typical f.m. frequencies (about 100MHz) and it may then be possible to construct an f.m. receiver without the use of a frequency converter.

SENSITIVITY

Although phase locked loops can operate with a signal of just under a millivolt, some problems do arise if one wishes to operate them from an input of about $10\mu V$ (such as one may obtain from a receiving aerial at a considerable distance from the transmitter). It is generally better to amplify the signal before feeding it to the phase locked loop. Some tuned circuits are usually included in the system before the phase locked loop in order to reject signals which could produce spurious responses.

The phase locked loop may be employed to demodulate both narrow and wide band f.m. signals. It offers greater linearity than can be obtained by any other means and is therefore especially suitable for high fidelity systems.

At the other extreme, phase locked loops are employed in communications receivers where they are excellent for recovering weak signals in a noisy region of the radio spectrum.

FREQUENCY SYNTHESISERS

Frequency synthesisers are required for use in radio communications networks which will generate any one of a large number of frequencies with great accuracy at the touch of a button. For example, an aircraft navigation/communication band synthesiser may be required to produce frequencies in the range 108 to 136MHz at 50kHz intervals so as to cover each of the 560 channels.

Such frequency synthesisers can be designed using banks of quartz crystals and tuned circuits. However, it is normally more economic to employ a single quartz crystal to provide one accurate reference frequency and to obtain the other frequencies from it by the use of a phase locked loop together with a digital pulse frequency dividing circuit. This system has the additional advantage that fewer spurious frequencies are generated.

The basic system used in a phase locked loop frequency divider is illustrated by the block diagram of Fig. 2.5. The reference frequency is generated by a very stable quartz crystal controlled oscillator which is kept in a special constant temperature oven. If the frequency of this oscillator is 100kHz, the pulses may be divided in frequency to produce, perhaps, a 1kHz signal. The latter is fed as the reference signal to the input of the phase locked loop of Fig. 2.5.

DIVIDER CIRCUIT

The main difference between the frequency synthesiser circuit of Fig. 2.5 and the simple phase locked loop is the inclusion of a frequency divider circuit

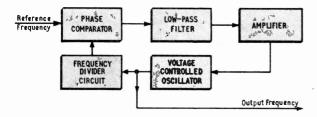


Fig. 2.5. The building blocks of a simple frequency synthesiser using a phase locked loop

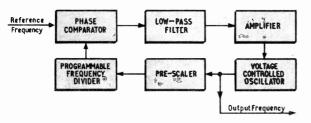


Fig. 2.6. In this type of frequency synthesiser a fast divider (the pre-scaler) is used to divide the pulse frequency.

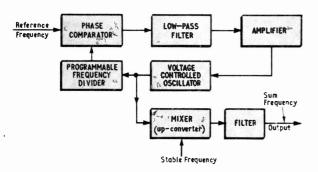


Fig. 2.7. Because phase locked loops can handle only relatively low frequencies, a mixer can be used to multiply the voltage controlled oscillator frequency.

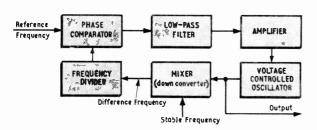


Fig. 2.8. In this type of frequency synthesiser a mixer is used to reduce the high voltage controlled oscillator frequency for the divider stage

between the voltage controlled oscillator and the phase comparator. This enables the voltage controlled oscillator to operate at a much higher frequency than the input reference signal.

If the reference signal has a frequency of 1kHz and the frequency divider is set so that it divides by, say, 569, the phase locked loop will come into lock when the voltage controlled oscillator frequency is 569kHz.

This frequency is divided so that it produces the same frequency as the input reference signal. In order to obtain any other output frequency from the voltage controlled oscillator, it is only necessary to change the factor by which this frequency is divided before it is fed to the phase comparator.

All frequencies generated in this way have the same stability as the reference frequency.

HARMONIC LOCKING

Phase locked loops will lock to a multiple of the input frequency, but the locking range decreases as higher and higher harmonics are used, since the amplitude of such harmonics is low.

Harmonic locking can be used in frequency synthesisers, but in general one should not attempt to lock a phase locked loop to a harmonic higher than about the tenth.

HIGHER FREQUENCIES

Digital frequency synthesisers are often required which will provide outputs of the order of some hundreds of megahertz. Voltage controlled oscillators can generate such frequencies, but programmable counters which can operate at above about 25MHz are very expensive.

It may therefore be more economic to operate the voltage controlled oscillator at a lower frequency and to multiply this frequency by a number of cascaded frequency multiplier stages. However, such a system has the disadvantage that tuned circuits are required in the multiplier stages and these must be re-tuned whenever the frequency is changed.

A more convenient system is shown in Fig. 2.6. The output from the voltage controlled oscillator is divided in frequency by a pre-scaler before being passed to a programmable frequency divider.

Other techniques for obtaining high frequency outputs include mixing the output from the voltage controlled oscillator with a stable frequency. The sum frequency is the required output, as shown in Fig. 2.7.

In yet another system, the output from the voltage controlled oscillator is mixed with a stable reference frequency and the difference frequency output is used to operate the programmable frequency divider. This type of system is illustrated in Fig. 2.8.

If a signal from a frequency synthesiser is fed to the mixer stage of a radio receiver together with the input signal, one can select the signal to be received merely by choosing the ratio by which the voltage controlled oscillator signal is divided. The drift in the tuning of such a system is quite negligible, since the reference frequency is crystal controlled.

Next month this series will continue with a review of the types of phase locked loops available from various manufacturers together with practical examples of the circuits in which they can be used.

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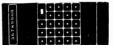
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A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.

This is YOUR page and any idea published will be awarded payment according to its merits.

PSU PROTECTION

When using a regulated PSU to reduce a supply voltage there is always the danger of component failure in the PSU and consequent damage to the equipment. A fuse will protect equipment when an excess of current is drawn but might be too slow to cope with overvoltage conditions.

In Fig. 1 the values shown suit a car battery supply being used to drive a battery tape recorder at 7.5V. The load requirement is 300mA and the trip voltage was set to 8V to protect the recorder in the event of regulator fault.

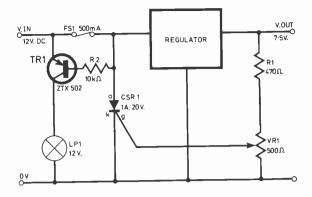


Fig. 1. Power supply protection circuit

R1 and VR1 form a potential divider which samples the output voltage as set by adjustment of VR1. The CSR1 is selected to carry at least twice the fuse rating. The full supply voltage is connected to the input of the regulator.

Transistor TR1 is held biased off by R2 and CSR1 so that the lamp LP1 is held off. If the output voltage rises above a set trip value then CSR1 will conduct, FS1 will blow and TR1 will be supplied with base current via R2 and the lamp will light up.

J. Sadler, Cheshire.

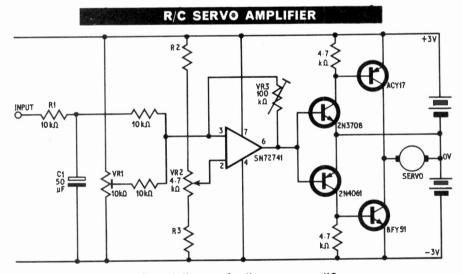


Fig. 1. Circuit diagram for the servo amplifier

DIGITAL proportional decoder was published in the series $Logical\ Radio\ Control$ (no issues available) and I have developed the following servo amplifier to go with such a device, see Fig. 1. Using a 741 operational amplifier for compactness, the circuit accepts the positive-going pulses, of variable length, from the decoder. These are integrated by RI and CI to give a d.c. which, using my decoder, varies from +0.2 to +1.2V.

Potentiometer VR1 is adjusted to give a value equivalent to minus the centre voltage at its wiper. In my case this is -0.7V. The 741 is used as a summation amplifier.

If VR3 is set at 10k\O so that the gain of the 741

is unity, then the output will be 1.2-0.7V = +0.5V at one transmitter control extreme and 0.2-0.7V = -0.5V at the other. In fact, the gain is set so that the output swings $\pm 3V$.

The transistors form a current amplifier to drive the servomotor and feedback from the motor is provided via the potentiometer VR2 which is ganged to the motor. The deadband of the servo may be controlled by varying VR3 and when the optimum value is found replacing with a fixed resistor.

The values of R2 and R3 are found by experiment. P. Skan,

Sutton Coldfold

Sutton Coldfield.

CAR LIGHTS MONITOR

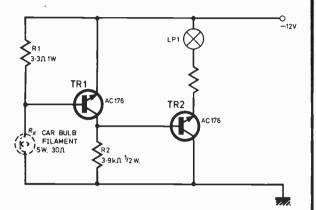


Fig. 1. Circuit diagram of the car lights monitor

PERHAPS the enclosed circuit of a car light monitor would be of interest to some of your readers. The circuit, Fig. 1, was designed to indicate the presence of a bulb failure in either the two front side lights or the two rear lights (the number plate light could also be included).

The 3.3Ω resistor R1 makes no visible difference to the bulbs light output as only approximately 1V is dropped across it. When the bulb filament R_x goes open circuit there is no forward bias to TR1 and its collector voltage rises and turns TR2 hard on, illuminating the indicator bulb LP1.

The circuit as shown will operate on positive earth cars; for negative earth cars pnp transistors, such as the AC128, should be used. The connecting details are shown in Fig. 2.

A wide variety of bulbs can be used in the LP1 position as long as the ratings of TR2 are not exceeded. The prototype used 6V 0·1A bulbs with 33Ω ballast resistors.

I. K. Staley, Mansfield, Notts.

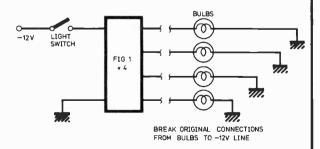
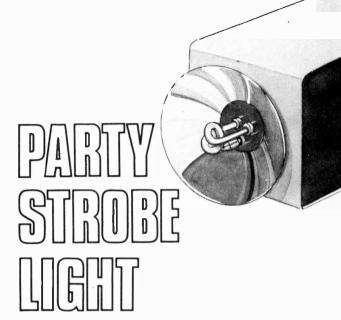


Fig. 2. Showing how the circuit is connected into the car wiring. The circuit in Fig. 1 is repeated for each light to be monitored

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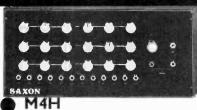
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4 high Z, 4 low Z inputs, 4 sets of controls. Case 10" × 8" and only 2\frac{1}{2} in deep. Carr. paid.

M6HL

12 inputs (6 high Z, 6 low Z) 18" × 8" × 2\frac{1}{2}". Carr. paid £27.50 + V.A.T.

Channel section modules, for building your own: gain—16 × (24dB). Tone controls—18dB swing. Carr. paid.

£3.50 + v.a.t.

CONTROL UNITS

Mono (as shown). **£6.50** Carr. 20p. For 9 v. battery operation. As stereo model, less mic.

Two decks, and full headphone monitoring. The unit is mains operated and measures 174in x 3in x 4in deep and is finished with a smart white on black facia. The controls are: Left/Right deck fader, volume, bass, treble, Headphone Selector and volume, Microphone volume, bass treble, mains on/off.
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• 160 watts version with power supply (Carr. 50p) £27.90

120 W HEAVY DUTY MODULE

Rugged class A driver stage. This module will run from all our mixers, etc., and most other makes. Delivers 120W into an eight ohm load and employs 4 TO3 can (115W) output transistors.

SPECIFICATION
Power output, 120W into 8 ohms
Freq. response, 20–20,000Hz ± 2d8
Input sensitivity, 200mV into 10K
Construction, Fibreglass board
Size, 8in x 4in x 4in (5in with supply)
Low distortion parallel push-pull output stage.

Another Saxon winner!



SOUND AND LIGHT UNITS

Our popular 3 channel model handles up to 3kW (3,000W) of lighting and incorporates versatile sound control arrangement to enable professional standards to be achieved. Both units are excellent examples of Saxon quality and value.

3 CHANNEL UNIT

Includes bass, middle and treble as well as master controls, 2 amplifier sockets eliminate need for split leads. Up to 3kW lighting load. Smartly finished steel case. £19.75 Carr. 30p.

SINGLE CHANNEL UNIT

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SAXON 100 £48-50 carr. free

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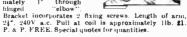
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W

NORTH AMERICAN SEMICONDUCTOR COMPANY LIMITED

TRANSI	STOR	S										٦
2N696	0 - 150	2N2219A	0 - 270	2N3108	0 - 540	2N3638	0 - 120	2N4135	1.500	2N5132	0-080	L
2N697	0 - 160	2N2221	0 - 190	2N3109	0 - 540	2N3638A	0 - 130	2N4140	0.270	2N5133	0.080	1
2N699	0 - 340	2N2221A	0 - 230	2N3110	0 - 540	2N3639	0 - 200	2N4141	0 - 290	2N5134	0.090	١N
2N706	0 - 100	2N2222	0 - 190	2N3117	0.850	2N3640	0 - 200	2N4142	0.310	2N5135	0.090	I١
2N708	0 - 145	2N2222A	0 - 245	2N3120	0 - 540	2N3645	0:150	2N4143	0 - 320	2N5137	0 - 100	١N
2N718	0 - 230	2N2243	0.930	2N3121	0 - 580	2N3642	0 - 150	2N4208	1-870	2N5138	0.090	I١
2N722	0.680	2N2243A	0.930	2N3133	0 - 260	2N3643	0 - 150	2N4209	2.000	2N5139	0.090	N
2N744	0 - 180	2N2270	0 - 390	2N3134	0 - 260	2N3644	0 - 160	2N4227	0 - 300	2N5140	0 - 170	N
2N753	0.180	2N2297	0 - 400	2N3135	0 - 260	2N3645	0 - 160	2N4228 2N4248	0.370	2N5141 2N5142	0 - 165	1.
2N760	0 - 270	2N2303	1 - 330	2N3136 2N3209	0.605	2N3646 2N3665	0.800	2N4246 2N4249	0 - 110	2N5142 2N5143	0 - 105	I.
2N760A	0.270	2N2368 2N2369	0 · 150 0 · 155	2N3248	1-690	2N3666	0.950	2N4249 2N4250	0.150	2N5910	0 - 270	l N
2N834 2N914	0 - 255	2N2369A	0 - 170	2N3246 2N3249	2.020	2N3691	0-950	2N4250 2N4257	0.175	BC107	0.130	I١
2N914 2N915	0 - 380	2N2483	0 - 210	2N3250	0-300	2N3692	0.150	2N4257A	0 - 230	BC108	0.420	1
2N915	0.300	2N2484	0 - 245	2N3250A	0 - 440	2N3693	0.150	2N4258	0 - 245	BC109	0-140	I'N
2N917	0.330	2N2509	0.710	2N3251	0.560	2N3694	0 - 150	2N4258A	0.275	BC177	0 - 180	IN
2N918	0 - 330	2N2510	0 - 710	2N3251A	0.620	2N3724	0 - 420	2N4274	0-135	BC178	0 - 180	l N
2N929	0 - 165	2N2511	1 - 130	2N3252	0-680	2N3724A	1 - 100	2N4275	0 - 135	BC179	0 - 180	LÑ
2N929A	0.740	2N2586	1.080	2N3253	0.740	2N3725	0 - 425	2N4313	0.250	BC261A	0 - 340	ΙÑ
2N930	0 - 190	2N2604	1 - 300	2N3299	0.340	2N3725A	1 - 090	2N4354	0-210	BC261B	0.340	١ĸ
2N930A	0-680	2N2605	0 - 840	2N3300	0 - 405	2N3734	1 - 170	2N4355	0.240	BC262A	0-360	IN
2N956	0.320	2N2657	3.070	2N3301	0 - 285	2N3735	2 - 700	2N4356	0 - 240	BC262B	0 - 360	١N
2N980	2 - 210	2N2658	4 - 500	2N3302	0 - 420	2N3903	0.165	2N4400	0-190	BC263A	0 - 380	1
2N995	1-080	2N2890	3 - 800	2N3304	0.700	2N3904	0 - 160	2N4401	0 - 240	BC263B	0-380	١.
2N1131	0 · 235	2N2891	3 - 500	2N3439	0.600	2N3905	0 - 175	2N4402	0 - 190	BCY70	0 - 155	11
2N1132	0-235	2N2894	0 · 320	2N3440	0 - 460	2N3906	0 - 165	2N4403	0.240	BCY71	0 - 220	1
2N1420	0 - 215	2N2894A	0.885	2N3444	0.830	2N3945	0 - 500	2N4916	0 110	BCY72	0 - 120	١.
2N1613	0 - 180	2N2904	0 - 225	2N3502	0.790	2N3946	1 - 200	2N4917	0 - 160	BFS92	0.655	1
2N 1711	0 - 220	2N2904A 2N2905	0.233	2N3503 2N3504	1 - 150	2N3947 2N3962	1·370 0·380	2N4943 2N4964	3 · 000 0 · 170	BFS93 BFX29	0-740	In
2N 1893	0 - 270	2N2905A	0.300	2N3504 2N3505	0.790	2N3963	0.455	2N4965	0.170	BFX30	0.330	Iъ
2N1990 2N2017	0 - 290	2N2905A	0 - 230	2N3545	4-500	2N3963 2N3964	0-455	2N4965 2N4966	0.170	BFX84	0.330	١ĸ
2N2102	0.375	2N2906A	0-245	2N3546	1 320	2N3965	0.760	2N4967	0.170	BFX85	0-280	I۸
2N2192	0 - 405	2N2907	0.210	2N3547	1.750	2N4030	0 - 465	2N4968	0.170	BFX86	0-230	I١
2N2192A	0 - 405	2N2907A	0-220	2N3548	1 900	2N4031	0.510	2N4969	0 - 170	BFX87	0.230	I١
2N2193	0.470	2N3011	0 - 180	2N3549	2 100	2N4032	0 - 530	2N4970	0 - 190	BFX88	0-210	I١
2N2193A	0.570	2N3012	0 - 530	2N3550	2 - 500	2N4033	0.670	2N4971	0.340	BFY50	0 - 190	P
2N2194	0 - 495	2N3015	0 - 405	2N3563	0 - 130	2N4036	0 - 480	2N4972	D - 360	BFY51	0 - 160	١٨
2N2194A	0-530	2N3019	0 - 540	2N3564	0:150	2N 4037	0 - 400	2N5055	0 - 200	BFY52	0-190	I١
2N2195	0 - 420	2N3020	0 - 500	2N3565	0 - 110	2N4121	0 - 135	2N5056	2.000	BSY38	0.170	Ľ
2N2195A	0.470	2N3053	0 - 180	2N3566	0 - 130	2N4122	0 - 150	2N5057	2 - 270	BSY39	0 - 170	10
2N2217	0.275	2N3054	0 - 450	2N3567	0 - 120	2N4123	0.145	2N5127	0.110	BSY51	0 - 200	1.
2N2218	0 - 215	2N3072	0 - 500	2N3568	0 - 170	2N4124	0 - 145	2N5128	0 - 140	BSY52	0 - 200	10
2N2218A	0 - 245	2N3073	0 - 455	2N3569	0 - 150	2N4125	0 - 160	2N5129	0.090	BSY53	0 - 200	1"
2N2219	0 - 240	2N3107	0 - 540	2N3576	1 - 229	2N4134	1 - 500	2N5130	0.080	BSY54	0 - 260	1
								2N5131	0 - 100	BSY95A	0 - 130	11

Available from THE RADIO SHOP

16 CHERRY LANE, BRISTOL BS1 3NG Tel. Bristol 421196. Open Mon.-Sat. 9.00 a:m.-5 30 p.m.

•	· CO: I	711	C11 11	10				
	THYRIS1	ORS						
3	Type No.	Price	(T (RMS)	DROM	NAS1640 NAS4510	0 - 342 0 - 350	1-6A 4-5A	400 100
ı١	NAS106F	D · 250	4A	50	NAS4520	0 - 400	4-5A	200
D	NAS106A	0 - 310	4A	100	NAS4540	0 - 640	4 · 5A	400
0	NAS106B	0 - 360	4A	200	NAS6510	0 - 390	6 · 5A	100
9	NAS106D	D - 580	4A	400	NAS6520	0 - 440	6-5A	200
3	NAS206F	0 - 283	6A	50	NAS6540	0 · 700	6-5A	400
0	NAS206A	0 - 349	6A	100	NAS8510	0 - 430	8 · 5A	100
5	NAS206B	0 - 405	6A	200	NAS8520	0 - 460	8-5A	200
5	NAS206D	0 - 653	6A	400	NAS8540	0 - 770	8 5A	400
5	NAS306F	0 - 313	8A	50	NAS10010	0 - 480	10A	100
Ð	NAS306A	0 - 389	8A	100	NAS10020	0 - 520	10A	.200
3	NAS306B	D-450	8A	200	NAS10040	0 - 840	10A	400
•	NAS306D	0 - 725	8A	400				
9	NAS406F	0.350	10A	50	l .			
0	NAS406A	0 - 434	10A	100				
0	NAS406B	0 - 504	10A	200	EPOXY 4		ND 6 AI	MP
	MACADED	0.840	10.4	400	DDIDGE	DECTIE	IEDC	

	NAS406A NAS406B NAS406D	0 - 434 0 - 504 0 - 812	10A 10A 10A	100 200 400			AND 6	АМР
	NAS806F NAS806A NAS806B NAS806D	0 · 438 0 · 543 0 · 630 1 · 015	16A 16A 16A 16A	50 100 200 400	7ype No. N7001 N7012 N7013 N7014 N7015	1F 4A 4A 4A 4A	VRM 50PIV 100PIV 200PIV 400PIV 600PIV	Price 0 · 320 0 · 360 0 · 430 0 · 590 0 · 755
	TRIACS V	S			N7021 N7022	6A 6A	50PIV 100PIV	0-360 0-430
-	Type No. NASQ1610	Frice 0-310	1-6A	DR OM 100	N7023 N7024 N7025	6A 6A 6A	200PIV 400PIV 600PIV	0 · 540 0 · 650 0 · 810
	NASQ1620 NASQ1640 NASQ4510	0 - 330 0 - 360 0 - 390	1-6A 1-6A 4-5A	200 400 100			PSULATE	D
	NASQ4520 NASQ4540 NASQ6510	0 - 440 0 - 710 0 - 420	4-5A 4-5A 6-5A	400 100	PLAST RECTII		50PIV	D · 350
	NASQ6520 NASQ6540 NASQ8510 NASQ8520	0 - 480 0 - 760 0 - 460 0 - 500	6·5A 6·5A 8·5A 8·5A	200 400 100 200	N6002 N6003 N6004 N6005	2A 2A 2A 2A	100PIV 200PIV 400PIV 500PIV	0 · 380 0 · 400 0 · 430 0 · 500
	NASQ8540 NASQ10010 NASQ10020 NASQ10040	0 · 830 0 · 500 0 · 550 0 · 880	8 · 5A 10A 10A 10A	400 100 200 400			CTIFIER	0.300
					1N4001			0.060

| NAS | 10020 | 0.550 | 10A | 200 | NAS | 100 | NAS | 1000 | NAS | 100



MAINS MOTOR

Precision made—as used in record decks and tape recor-ders—ideal also for extractor fan, blower, heaters, etc. New and perfect. Snip at 72p. Postage 20p for first one then 10p for each one ordered

MUSIC ON TAPE

A further buy enables us to offer these at an even lower price—namely 65p each or 5 for \$2.50, Send for list of titles. We can't repeat when sold out.

BALANCED ARMATURE UNIT

500 ohm, operates as speaker or micro-



Fits in place of cigarette lighter. Useful method for making a quick connection into the car electrical 88p each or 10 for \$3.42. 12 VOLT IL AMP

This comprises double-wound 230/240V mains transformer with full wave rectifier and 2000 mF smoothing. Price 22.20, plus 20p post & packing.

42.20, plus 20p post & packing.

Heavy Duty Mains Fower Pack. Output voltage adjustable from 15-40V in steps—maximum load 250W—that is from 6 amp at 40V to 15amp at 15V. This really is a high power heavy duty unit with dozens of workshop uses. Output voltage adjustment is very quick—simply interchange push on leads. Silicon rectifiers and smoothing by 3,000mF. Price 26-33 plus 65p post.



BAKELITE INSTRUMENT CASE

Size approx. 64"×31"×2"
deep with brass inserts in four corners and bakelite panel. This is a very strong case suitable to house instruments and special rigs, etc. Price. 50p each.

Paxolin lid 11p extra.



MINIATURE

WAFER SWITCHES
2 pole, 2 way—4 pole, 2 way—3 pole, 3 way—4 pole, 3 way—2 pole, 4 way—2 pole, 4 way—1 pole 6 way—1 pole, 12 way. All at 25p each. each

CONNECTING WIRE

7-0076 Copper conductors. 500 metre druns available in the following colours: Red/Brown, Yellow. Green/Grey, Blue/Green, Red/Grange. Green/White, Grey/White, Blue/Grange, Brown/Red, Brown/White, Red/Grey, Blue/Grey, Blue/Frown. Price \$2.20 per drunn plus 40p post. State alternative colours.

Screened Cable. 70076 core suitable for pick up or mike lead or for inter-connecting amplifiers. 16 metres, 33p.

DOOR SWITCH

As fitted to refrigerators, etc. Switches off as door closes. White 17p each. Black 15p each.

MICRO SWITCH

5 amp changeover contacts, 11p each. 10 for 99p, 15 amp on/off. Model 15p each. Changeover 19p each.

U.V. LIGHTING

U.V. LIGHTING
Useful for flaw detection in metals and for looking for water marks, etc., also for fitting over tropical fish tanks—African violets and other indoor plants which must have U.V. for healthy growth. The outfit comprises of a 12 in U.V. tube—choke—starter holder and tube ends. Price \$2.20 plus 20n noat 20p post

80 WATT TUBULAR ELEMENTS Brass encased with beaded flex ends. Standard replacements in most absorbtion type refrigerators but also has dozens of other uses. 83p each.



SPIT MOTOR 200-250V Induction

200-250V Induction Motor, driving a Carter gear box with 1 jin. of output drive shaft running at 5 revs per minute. Intended for roasting chickens also suitable for driving models—windmills, coloured disc lighting effect, etc., etc., etc., 42-04 plus 20p post and insurance.

SWITCH TRIGGER MATS
So thin is undetectable under carpet but will switch on with slightest pressure. For burglar alarms, shop doors, etc.
24in x 18in £1-82.
13in x 10in £1-21.

ROCKER SWITCH 13 amp self-fixing into an oblong hole. Size approximately 1" x 2" 9p each, 10 for 82p.

KETTLE ELEMENTS

Made by the famous A.E.I. Co. Complete with washers and combined fixing ring and plug shroud. Normal 2 round pin and flat pin earth connection and overload reset push button, 2 Models—1\$\frac{1}{2}\times \text{misslar}\text{misslar}\text{bulk





MULLARD AUDIO AMPLIFIERS
All in module form, each ready buin complete with heat
sinks and connection tags, data supplied. Model 1153 500mW power output 72p. Model 1172 750mW power output 94p. Model EP9000 4 watt power output \$1.86. Unilex DIY STEREO. Complete \$11.30.

CENTRIFUGAL BLOWER

CENTRIPUGAL BLOWER
Miniature nains driven blower centrifugal type blower
unit by Woods. Powerful but specially built for quiet
running—driven by cushioned induction motor with
specially built low noise bearings. Overall size 4½" × 4½"
* 4". When mounted by flange, air is blown into the
equipment but to suck air out, mount it from centre
using clamp. Ideal for cooling electrical equipment or
fitting into a cooker hood, film drying cabinet or for
removing flux smoke when soldering etc. etc. A real
bargain at 22-05.

MIGHTY MIDGET
Probably the tiniest possible radio, as described in Practical Wireless, 4 January
73. All electronic parts \$2:20 post paid.

TAPE PLAYBACK UNITS

Mains operated. Made by Reditune the famous "music in the background people". These are complete units for playing tape at standard speed (347). These have a superior motor driven fly wheel to control the tape through the capstan and also an even equally useful valve amplifier with EL84 output. In a steel case with carrying handle. Tested and in good working order 26.50 or some in need of repair but complete 25.50. Cassettes of tape pre-recorded 21.10 (please state type of music required). 75p carriage up to 200 miles then 50p per 100 miles extra.



6 MAINS TRANSFORMERS

Our Rel: MTM1 27V at 8A. Upright mounting fully shrouded—flying leads—fully tapped primary. Price 28-30.

Our Rel: MTM2. 12V at 1A. Upright mounting with fixing lugs, tag connections. 240V primary—12V 1A secondary. Price 28p.

Our Rel: MTM3. 6-3V 2A upright mounting with fixing lugs, tag connections, 240V primary 6-3V 2A secondary. Price 77p.

Our Rel: MTM4. 18V—1A with thermal cut-out, upright mounting with fixing lugs—tag connections. Primary 240V—secondary 18V at 1A. Price 88p.

Mains Isolation, 350W earth shielded—flex leads—upright mounting lugs for fixing terminals for output. Price 26-50 each.

fixing terminals for output. Price \$6.50 each.

Mainr Transformer Bargain. Standard mains 240V input. Secondary 2-4V 9A intermittent 5A continuous. Price 55p.

DISTRIBUTION PANELS

Just what you need for work bench or lab. 4×13 amp sockets in metal box to take standard 13 amp fused plugs and on/off switch with neon warning light. Supplied complete with 7 feet of heavy cable. Wired up ready to work, £2.50 plus 20p post and insurance.



24-HOUR TIME SWITCH

Made by Smiths, these are A.C. mains operated, NOT CLOCKWORN. Ideal for mounting on rack or shelf or can be built into box with 13A socket. Two completely adjustable time periods per 24 hours, 5A changeover contacts will switch circuit on or 5A during these periods. 22-75 post and ins. 23p. Additional time contacts 55p pair.

THYRISTOR LIGHT DIMMER

For any lamp up to lkw. Mounted on switch plate to fit in place of standard switch. Virtually no radio interference. Price \$2.95, plus 20p post and insurance. Industrial mode! 5A \$3.30. Not on plate.

- THIS MONTH'S SNIP-

WALL THERMOSTATS

Made by the famous Smiths Instrument Co. called Colourstat. Wall mounting and in a handsome plastic case (cream and beige). Adjustable by silder (lockable) and may be set to control temperatures from around freezing through to 50°C. The silde panel is engraved and indicates "frost, "warm", "very warm", etc. The thermostat will control heaters, etc., up to 15A at normal mains voltage and is ideal for living room, bed room and greenhouse, etc. Price 21-65. Don't miss this.

THE TWENTYLITE

Fluorescent lighting units with polyester choke and finished white enamel. 2ft. model, ideal kitchen, bedroom, hallway, porch, lift, etc., with tube. Assembled ready to install. Price \$2.20 + 40p P. & P.



10 AMP. DIMMER/CONTROLLER

For the control of lighting of stage or studio or portable equipment in workshops, etc. This has socket outlets each controlled by 5 AMP Solid State Regulator. Also fitted with master switch fuse and neon indicator and terminating with 6 feet of flex. Overall length 17in. 3½in. x 1½in. deep. £7:50 plus 25p P. & P.

TANGENTIAL HEATER UNIT
This heater unit is the very latest type, most efficient, and quiet running. Is as fitted in Hoover emcient, and quietrunning. Is as fitted in Hoover and blower heaters costing £15 and more. We have a few only. Comprises motor, impeller, 2kW element and 1kW element allowing switching 1, 2 and 3kW and with thermal safety cut-out. Can be fitted into any netal line case or cabinet. Only needs control switch. £3.85, 2kW Model as above except 2kW £2.75, Don't miss this. Control Switch 44p, P. a P. 40p.



Voltage Changing Transformers made by Parmeko. Upright mounting, fully shrouded and with terminal blocks for input and output. For changing mains voltage, ideal for working low voltage equipment from 230/240 mains and for increasing voltage due to losses in long leads. Voltage up or down between 190-250V 250W. Price \$1.65 with 300 not set. plus 30p post, etc.

Door Opening or Platform Rotating Motors. Very powerful motors estimated rating at \{ \(\hbar \)_P. Reversible with gearbox and "\" belt drive wheel of 7 in diameter. Price of motor which weighs approx. 15lb is \(\frac{\pmathbf{1}}{2} \) Mot. These are ex-equipment but guaranteed for six months.

Gas and Smoke detector/Sensor—ref. GD1—recently referred to as the electronic nose—22 each; circuit of smoke detector slarm included. Constructional details appeared in September's Practical Wireless, so we can supply the SL03D at \$1.75, the ZN414 at \$1.85, also most other items, send for PW10 radio parts list.

Spring Return Water Switch. As used in inter-com and other similar equipment a two wafer 6 pole 3 way switch, spring return from centre position when turned clockwise and permanent off or on when turned anti-clockwise. Price off or or 55p each.

A.C. Buzzers. 12V. Fix these into a box which A.C. Buzzers. 12V. Fix these into a box which will resonate and they will give a loud piercing note. Suitable for alarms or signal. 33p each. Push-on Tag Connectors. These are being increasingly used on care and domestic appliances. 2 sizes prevail, the most popular being for jin tag. We offer these at attractive prices. 10 for 10p; 30 for 40p; 100 for 70p; 1,000 for \$5.

4 Minute Timed On Switch by Smiths. Our ref. T8 AU1. This is for processing photographic plates, etc. Has a scale graduated 0-4min. Turn the polnter knob, and set it at the desired time—switch will stay closed until pointer knob returns to zero. Clockwork motor, so can be independent of mains if desired, but the switch itself is double pole, rated 250V, 15Amps. Price 21-65. £1.65

One Hour Timed On Switch by Smiths. Our ref. TS AU2. As TS AU1 except that the complete rotation of the pointer knob lasts for 1hr. Price £1-10 each.

CONTROL DRILL **SPEEDS**

CONTROLLER NEW IKW MODEL Electronically changes

speed from ately 10 approxispeed from approxi-mately 10 revs. to maximum. Full power at all speeds by finger-tip control. Kit includes all parts, case, everything and full instruc-tions. 21-95. 139 post and p. Made up model also available. 22-95. revs

MAINS TRANSISTOR POWER

Designed to operate transistor sets and amplivolts for up Designed to operate transistor sets and ampli-fers. Adjustable output 6v., 9v., 12 volts for up to 500mA (class B working). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6, PP7, PP9, and others. Kit comprises: mains transformer rectilier, smoothing and load resistor condensers and instructions. Real anip at only 21-10. Pluz 20p postage.

Stereo Pre-amp Module. Muliard ref. LP 7182/2—transistors and all ancillary components mounted on PC board £1-32 with connection dig.



ISA ELECTRICAL PROGRAMMER

PROGRAMMER
Learn in your sleep:
Have radio playing and kettle boiling as you awake-switch on lights to ward off intruders have a warm house to come home to. All these and many other things you can do if you invest in an electrical programmer. Clock by famous maker with 15 amp, on/off switch. Switch-on time can be set anywhere to stay on up to 6 hours. Independent 60 minute memory jogger. A beautiful unit. Price \$2·15 + 20p p. & p. or with glass front chrome bezel 83p extra.

THERMAL CUTOUT

THERMAL CUTOUT

A miniature device 2in dis. on one screw fixing mount—can be used for motor overload protection, fire slarm, soldering iron, switch of, etc., etc. 15 amp contacts open with flame—radiant or conducted heat. 11p seach or 10 for 99p.

HONEYWELL PUSH BUTTON MICRO SWITCHES

As used in some vending machines, etc. Each switch is a changeover type rated at 15A 250V a.c. Intended for panel mounting with fixing nuts and single black push knob. Single changeover switch 28p, Double switch 42p, Triple switch 61p.

MACLAREN THERMOSTAT

Make and break 20A a.c. with the sensor probe coupled by 2 feet capillary covering range of 10-100°C—complete with large engraved control knob. Price 83p.

TERMS: 10% discount if ten of an item ordered, send postage where quoted-other items, post free if order for these items is £6.00, otherwise add 20p.

J. BULL (ELECTRICAL)

(Dept. P.E.), 7 Park Street, Croydon CRO IYD Callers to 102/3 Tamworth Road, Croydon

AUDIOTRONIC MODEL ATM.1

Top value 1,000 o.p.v. pocket multimeter.
Ranges: 0/10/50/250/1,000V. A.C. and D.C. D.C. Current 0-1mA/ 100mA.

100mA. Resistance: 0/150 K ohms. Decibels: -10 to +22dB. Slze: $90 \times 60 \times 28$ mm. Complete with test leads. **22.95.** Post 15p.



RUSSIAN 22 RANGE MULTIMETER Model U437 10,000 O.P.V. A first class versatile instrument manufactured in U.S.S.R. to the highest standards. Ranges: 23,10/50/220/500/1,000V d.c. 2-3/10/50/220/500/1,000V d.c. D.c. current 100wA/1/10/100mA/1A. Complete with batteries, test leads, instructions and sturdy steel carrying case. sturdy steel carrying case OUR PRICE \$4.95. Post 25p



MODEL 500 30,000 O.P.V with overload protection mirror scale 0/0.5/2.5/20/25/ 100/250/500/1,000V d.c. 100/250/300/1,000V d.c. 0/25/10/25/100/25/100/25/100/25/100/25/100/25/500/1,000V a.c. 0/50μA/5/50/500mA. 12 amp. d.c. 0/60/K/6 Meg./60 Meg. 60 Meg. 210-50. Post paid. Leather case £1.75.



0.00

U4312 MULTIMETER

U4312 MULTIMETER

Extremely sturdy instrument for general electrical use, of the study instrument for general electrical use, of the study of the s



100,000 O.P.V. Overload protection. Mirror scale, 03/0-6/1-2/1-5/8/6/12/30/60/ 120/800/600/1.200V. d.c. 1-5/3/6/12/30/60/150/300/ 600/1.200V. a.c. 15/30/A/3/6/30/6/150/300m. 15/30/A/3/6/30/6/150/300m. 6/12 amp. d.c. 2K/200F 2 Meg/20 Meg ohm -20 t +63dB. £14.95. Post 20p



Giant 6in mirror scale. 20,000 O.P.V. 20,000 O.P.V. 0/0·25/1/2·5/10/50/250/ 1,000/5,000V d.c. 0/2·5/10 /50/250/1,000/5,000V a.c. 0/50µA/1/10/100/500mA/

10 amp. d.c. 0/2K/200K/ 20 meg. -20 to +50dB. \$13.95. Post 35p.

MODEL C-2080 EN

370 WTR MULTIMETER

Features a.c. current ranges, 20,000 O.P.V. 0/P.5/2-5/10 / 50 / 250 / 500 / 1,000V d.c. 0/2-5/10/50/2-50/500/ 1,000V

0/50µA/1/10 / 100mA / 1 / 10

amp. d.c. 0/100mA/1/10 amp. a.c. 0/5K/50K/500K/5MEG/50MEG. -20+62dB. £17-50. Post 25p.

TMK LAB TESTER

100,000 O.P.V. 64in scale buzzer short circuit check. Sensitivity: 100,000 OPV d.c. 5/Voit a.c. D.c. volts: 0.5, 2.5, 10, 50, 250, 1,000 V.



10, 50, 550, 1,000 V. A.c. volts: 3, 10. 00 V. D.c. current: 10, 50, 250, 500, 1,000 V. D.c. current: 10, 100 μ A, 10, 1,00, 500 μ A, 2. 5, 10 A. Resistance: 1K, 16K, 100 K, 10 μ C, 100 meg. Peciphels: 10 to 1+49GB; Plastic case with carrying handle, size $74 \ln \times 64 \ln \times 34 \ln$ £19-95. P. & P. 290.

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Model S-100TR MULTIMETER/TRANSISTOR

TESTER. 100,000 O.P.V. mirror scale/overload protection. 0/0·12/0·6/3/12/30/120/ TESTER. mirror scale/overload protection. 9(0-12/0-6/31/2/30/120/ fo00 d.c. 9/6/30/120/fo00 a.c. 0/12/500/a/12/300mA/12 amp. d.c. 0/10 K/1 MEC/1/100 MEG. -20 to +50dB. 0-01-2 MFD. Transistor tester neasures Alpha, beta and Ico. Complete with batteries, instructions and leads. 214-95. Post 25p.



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Tests ICO and B. Pnp/npn. Operates from 9V battery. Complete with all instructions, etc. £3-95. Post 20p.



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Tests PNP or NPN transistors. Audio indication. Operates on two 1-5V batteries. Complete with all instructions, £4.50. Post 20p.



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Checks true a.c beta in/out. Checks 1cbo. Checks diodes in/out. Checks 8CR, etc. Beta H1 10-500 LO 2-00 1bco 0-5000µA. 220/240V a.c. operation. 217-50. Post 25p.



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High quality instrument to test Reverse Leak current and D.C. current. current and D.C. current.
Amplification factor of
NPN, PNP, transistors,
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4½" clear scale meter.
Operates from internal
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TE-65 VALVE VC
28 ranges. D.c. volts
1:5-1,500V. Resistance
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200/240V r a.c. operation. Complete with
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MULTITESTER
High sensitivity tester.
200,000 o.pv. Overluad
protection. Mirror scale.
Ranges: 0/0-06/0-3/3/30/
120/609/01/1,200V d.c., α/3/
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a.c. -20 tu +63dB. 0/
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7691109,5007,760 d.c. 07 750mV[1-5]8[7-5]15[30] 75]150/300/750V a.c. Automatic cut out. Supplied complete with test leads, manual and test certificates. 249. Post 50p.



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10 meg. input 10 ranges: 01/-003/-1/-3/1/3/13/10/30/100/30/N. R.M.S. 4cps.-1-2 Mc/s. Decibels —40 to +50dB. Supplied brand new complete with leads and instructions. Operation 230V. a.c. \$17-50. Carr. 25p.

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VOLTMETER
Battery-operated, 11
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Large 4\(\frac{1}{2}\)in mirror scale.
8\(\frac{1}{2}\)ic 0.3-1.20\(\frac{1}{2}\)in \(\frac{2}{2}\)in \(\frac{1}{2}\)in \(\frac{1}\)in \(\frac{1}{2}\)in \(\frac{1}\)in \(\frac{1}{2}\)in \(\frac

£17.50. P. & P. 20p.

KAMODEN HMG-500 INSULATION RE-SISTANCE TESTER

Range 0-1,000 Megohms, 500 Volt. Megohms, 500 Volt, Battery operated. Wide range clear meter 41" × 4". Complete with deluxe carrying case, batteries, instructions. £19-95. Post 30p.



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Transistorised. Operates
as Grid Dip, Oscillator,
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Frequency range 440kHz280MHz in 6 coils, 500µA
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Sin tube, Y sunp. Sensitivity
0-1V p-p/CM. Bandwidth
1-5cps-1-5MHz. Input
imp. 2 meg D 25pF X sunp.
sensitivity 0-9V. p-p/CM.
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SMHz Pass Band. Separate Y1 and Y2
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ARF-300 AF/RF SIGNAL GENERATOR SIGNAL GENERATUR.
All transistorised, compact, fully portable. AF
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Frequency 0-200kHz. 0-200k Hz. Attenuator0-111dB.



O'ldB step.
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Size 180mm × 90mm × 55mm.
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Square 18-50,000 Hz. Output max. +10dB(10 K ohms) Operation internal batteries. Attractive 2-tone case 7 in × 5 in × 2 in. Price £17-50, Carr. 17p.



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	1	amp.			£7-00
,	2.5	amp.			£8-05
		amp.			£11.75
		amp.			£15-90
		amp.	÷		£22·50
		amp.			£28·60
		amp.			£49-00
		amp.			£58-00
	40	amp.			£82.50
2					

MODEL S-260B

3-2000						
Panel mounting						
lamp £7.00						
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Carriage and						
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AUTO T	RANSE	ORMERS	
0/115/230V. Step			Fully
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80 W	£2.85	P. & P. 18p	
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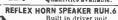
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80 W	£2-85	P. & P. 18p
150W	£3-00	P. & P. 18p
300W	£4-00	P. & P. 23p
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Brand new. 3 sets of changeover contacts at 5 amp. rating. 40p each. Post 10p (100 lots £30) Quantities available.





Built in driver unit.
Impedance 16 ohm. Power rating 10W. Response 380-7,000Hz. Approx. size 6in x 6in. Weatherproof and shock proof.
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Approx. 7-1 ratio planetary drive vernier
dials. Log scale
0-180 degrees. Blank
scales 1 to 5. Scale
width 4lin. Dial size
in x 3in. Overall size
7½ in x 4kin x 1½ in
and coupling, 1in dia.
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Over 200 ranges in stock-other ranges to order. Quantity discounts available. Send for fully illustrated brochure.

50mA 100mA

500mA

l amp.

15 amp

Type MR.85P. 4jin. x 4jin. Fronts

TYPE 8W.100	100m	m × 80mm Fronts
0µА 0-0-50µА 100µА	24-15 23-95 23-95	The state of the s
100-0-1 0 0µА. 100µА	£8-90 £3-70 £3-60	
0V d.c 0V d.c	28-60	5 amp. d.c
lamp.d.c		VU Meter 24-8

Type SD.880 82-5mm × 110mm Fronts | 10mA £3.10



	000mA	£3.1
	1 amp	
	5 amp	£3-1
	10 amp	
50μA £8-40	5V d.c. ,	£3-1
	10V d.c	
10 0 μA £3·35	20V d.c	.23-1
100-0-100μA . £3-35	50V d.c	23-1
200μA £3·80	300V d.c	£3-1
500μA 23-15		£3-3
1mA £3·10	300 V a.c	£3.3
5mA £3.10	VU Meter	£8.5

50m A 100mA

Type 8D.640 6	33-5 mm	× 85mm Fronts
50μA	£3.05	500mA £2-90
50-0-50μA	£3.05	1 amp £2-90
100μA	£3.00	
100-0-100μA .	£8.00	10 amp £2-90
200μA	£3-00	5V d.c £2-90
500μA	£2-95	20 V d.c £2-90
1mA	£2-90	50V d.c \$2-90
5mA	£2-90	300V d.c 22-90
10mA	£2.90	15V a.c £3-00
50m A	£2-90	300V a.c 28-00
100mA	£2.90	VU Meter £3-15

Type SD.460 46mn	× 59-5mm Fronts
50μA £2.8	0 1 amp £2-60
50-0-50μA £2-8	0 5 amp £2-60
100μA £2.7	
100-0-100μA . 22-7	5 5V d.c £2.60
200μA £2.7	0 10V d.c £2-60
500µA £2.5	5 20V d.c £2.80
1mA £2-6	
5mA £2.6	
10mA £2.6	153
50mA £2.6	
100mA £2-6) 300 Va.c £2.70
500mA £2-6	VU Meter £2-90



" SEW " **EDGEWISE METERS**

Type	PE.70.	3 17 21in. (/32in. × leep.	1 15	/82in. ×	
50-0-5	0μΑ	#3-60	1mA		£3.20	nie
100μ/	100μΑ	#3-50	300V a		£3.25	de

*MOVING IRON-

SMEAN ATTEN	CAT FORE IN
5-min	1
1	. "
m.	A
Δ.	1 .
SAIL 12	
13.65	
100000000000000000000000000000000000000	
- mm	Oldan.

·	See	50V d.c
50μA	24-40	150V d.c
50·0·50μA	24-25	300V d.c
100μΑ΄	24-25	15V a.c
100-0-100дА.	£4-05	300V a.c
200μA	24.05	8 Meter ImA.
500μA	£3.90	VU Meter
. Au 006-000c	£3.90	1 amp. a.c.* .
lmA	23-90	5 amp. a.c
1-0-1mA	£3.90	10 anip. a.c.*
imA	£3-90	20 amp. a.c.*
10mA	£8.90	30 anip. a.c.*

A O A1111-1	20.00	TO WILL WICE	40.90	1 1 0 - 11
5mA	£3-90	20 amp. a.c.*	£3.90	2miA
10mA	£8-90	30 anip. a.c.*	£8-90	5mA
1				10mA
				20m A
				50mA
Type MR.52P.	2]in. s	quare Fronts.		100m
30μA	£3.50	10V d.c	£2.50	150mi
50-0-50μA	£3.05		£2.50	
100μA	£3.00	50V d.c	42-50	! —
100-0-100μA .	£2.95	300 V d.c	£2.50	Type 1
500μA	£2-65	15V a.c	£2-60	50цА.
lmA	\$2-50	300V a.c	£2.60	50-0-5
5mA	£2.50	8 Meter 1mA.	£2-60	10044
10mA	£2.50	VU Meter	£3-60	100-0-
50mA	£2.50	l amp. a.c	£2.50	200µA
100mA	£2-50	5 amp. a.c.* .	£2·50	500µA
500mA	£2.50	10 amp. a.c.*	£2.50	500-0-
1 amp	£2-50	20 amp. a.c.*	£2·50	1mA
5 amp	£2.50	30 amp. a.c.*	£2.50	5mA

£3-00 £8-00	Type MR.65P.	8 in. ×	3in. Fronts	
£3.15	50μA	£3.70		£2.60
F0.10	30-0-50μA	£3-15	20V d.c	£2.60
	100μΑ	£8-15	50V d.c	£2-60
_	100-0-100μA .	£3-10	150V d.c	£2-60
8	200μA	£3-05	300V d.c	£2.60
£2-60	300μA	22.75	15V a.c.	£2-80
£2-60	500-0500μA .	£2-60	50V a.c	€2-80
£2-60	1mA	£2-60	150V a.c	£2-80
£2.60	5mA	£2-60	300V a.c	£2-80
£2-60	10mA	£2-60	500V a.c	42.80
£2.80	50mA	\$2.60	8 Meter ImA.	£2-85
£2-80	100mA	£2-60	VU Meter	23-70
£2:60	500mA	£2.60	50mA a.c.*	42-60
£2.70	1 amp	£2-60	100mA a.c.* .	22-60
£2.70	5 amp	£2-60	200mA a.c.* .	\$2.60
	10 amp	22-60	500mA a.c. ,	\$2.60
£2.90	15 amp	£2-60	l amp. a.c	£2-60
	20 amp	£2-60	5 amp. a.c	£2-60
	30 amp	£2.80	10 amp. a.c.*	\$2.60
	50 amp	£2-90	20 amp. a.c.*	\$2.60
_	5V d.c	22-60	30 anip. a.c.*	£2-60
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"SEW" EDUCATIONAL METERS



	24117
Type ED.107. Size overall 100mm × 90mm × 108mm.	10m
A new range of high quality moving coil instruments ideal for school experi-	50m 100r 500r
ments and other bench applications. Jin. mirror scale. The is easily accessible to nal working. Available nges:	50µ. 50-0 100µ 100-
90 110 1 c #5.95	500 _L

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ieter movement	is easily accessible to
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0иA £6-1	nges: 90 10V d.c
00μA £6.	0 20V d.c 25.95
0-0-50μA £6-4	10 50V d.c £5.95
tnA £5-1	95 300V d.c £5-95
-0-1mA #5-1	Dual range

Type MR.38P, 1 21/82in, square Fronts. 200mA ... \$2-85 300mA ... \$2-25

42.95



100μΑ..... 100-0-100μΑ .

200μΑ 500μΑ 500-0-500μΑ 1mA 1-0-1mA

\\$b	DUUMA	£2.20
•	750mA	42-25
	1 amp	42-25
4 _	2 amp	42-25
	5 amp	42-25
	10 amp	42.25
	3V d.c	£2-25
£2.55	10V d.c	42-25
£2.50	15V d.c	42-25
£2-45	20V d.c	£2-25
£2-40	50V d.c	42-25
£2-25	100 V d.c	22.25
22-25	150V d.c	£2-25
22-25	300 V d.c	22.25
42-25	500V d.c	42-25
£2-25	750V d.c	42-25
€2-25	15V a.c	42.35
£2-25	50V d.c	£2-80
€2.25	150V a.c	\$2.80
£2-25	300 V a.c	£2-80
€2-25	500V a.c	€2-80
22-25	8 Meter 1mA.	
£2-25	VU Meter	
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5mA ,	22-25	50V d.c	£2-80
10mA	22.25	150V a.c	42.80
20mA	£2-25	300 V a.c	£2-80
50mA	£2-25	500V a.c	€2-80
100mA	22-25	8 Meter 1mA.	42-30
150niA	£2-25	Vl Meter	£2-65
Type MR.45P.	lin. squ	are Fronts.	
50μA	\$2.70	5 amp	£2-40
50·0·50μA	£2.65	10V d.c	£2-40
100μΑ	£2-60	20 V d.c	£2-40
100-0-100µA .	£2-50	50V d.c	\$2.40
$200\mu A$	£2-50	300V d.c	£2-40
500μA	£2-45	15V d.c	€2-40
500-0-500μA .	£2-40	300 V d.e	22-40
1mA	#2-40	8 Meter 1mA.	£2.50
5mA	£2-40	VI Meter	£2.70
10mA	#2-40	1 amp. a.c.* .	£2-40
50mA	£2-40	5 amp. a.c	£2-40
100mA	£2·40	10 amp. a.c.*	42-40
500mA	42-40	20 amp. a.c.*	£2-40
1 amp	#2-40	30 amp. a.c.*	£2-40
		-	

"SEW" BAKELITE PANEL METERS

30 amp. 50 amp.

Type MR.65. Siin, square Fronts. l amp. 5 amp. 15 amp.



50μA...... 50-0-50μA...

100μΑ....

500μA..... 500-0-500μA.

1mA ... 1-0-1mA

5mA

10m A

50m A 100m

300 V d.c

micro-

	0 1 a.c	27. GO
	10V d.c	£2-60
	20V d.c	£2-60
	50V d.c	22.60
	150V d.c	\$2.60
	300 V d.c	£2.60
£4-60	50mV d.c	\$2.90
£3-55	100mV d.c	42-90
23.05	30 V a.c.*	\$2.65
£3-00	50V a.c.*	\$2.65
£3.00	150V a.c.*	£2-65
£2.70	300 V a.c.*	\$2.65
£2-60	500mA a.e	£2-60
£2-60	lamp.a.c.*.	£2-60
£2-60	5 anip. a.c. * ,	£2-60
22-60	10 amp. a.c.*	£2.60
£2-60	20 amp. a.c.*	\$2-60
£2-60	30 amp. a.c.*	£2-60
£2-60	50 аптр. а.с. *	£2-60
£2.60	VU Meter .	28-65

£2.60 £2.60 £2.60

£2-60

Type S-80 80mm squar	e Fronts
50µA £3-50	
50-0-50µА	11
100μΑ #3-40	11 11
100-0-100µA #3-30	II PA
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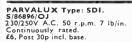
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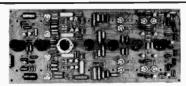
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PC7 avail. to order 43-20.

ELECTRONIC PIANO (PE Sep. 72/Jan. 73). PCB PRICES DOWN. Details in lists.

GEMINI STEREO AMPLIFIER

GEMINI STEREO AMPLIFIER

Maka-Sw's—with †W MO Rs, £10-90—with †W or †W CF Rs, £6:-90, PCB (3‡in × 10‡in) for sets with MO and †W Rs, also holds rotary or slider pots and Maka-Sw's, £1-90, Main Amp—Rs, Cs, Pots, £5-20, PCB (3‡in × 5in), £1-15. PSU—Rs, Cs, Pot, £3-70, PCB (2in × 4in), 65p.

GEMINI STEREO TUNER
(PE Apr./Jun. 72). Rs, Cs, Pots,
3-80. PCB as publ., £1-50.
MICROPHONE MIXER
(PE Apr. 69). 5/c's, Rs, Cs, Pots,
£2-60. PCB (3½in x 4½in) also holds
6 rotary or 4 slider pots, £1-10.

LOGICAL RADIO CONTROL AND MODEL SERVO CONTROL
(PE Feb./Jun. 72)—Details in lists—Hurry not many left!

PHONOSONICS

NEW LOW PRICES FOR MANY PCB's AND KITS



HI-FI TAPE LINK

(PE Mar./Apr. 73). S/c's, i.c.'s, Rs, Cs, Relay and pc-base, Pot Cores and pc-bases, Sw's, Pots Panel Lamp—Mono, £11-80; Stereo, £18-70. PSU', £2-50. PCB—Main Circuit (3½ in × 9in) (Stereo) also holds Sw, Rs, Cs, Presets and mounts on Sw's, 80p.

REVERBERATION UNIT

(PW Nov./Dec. 72). S/c's, Rs, Cs, Slider Pots, T/fmr, £6-80. PCB (2in × 11±in) also holds sliders, £1-20. 9in Spring Unit avail. to order, £3-75.

asso nolds sliders, £1-20. 9in Spring UPBRASONIC GUITAR
PRE-AMP
(PW Sept. 70). Incl. Mic P/A, 2Guitar P/A, Trem and Tone Controls,
Master Volume. S/Cs, Rs, Cs, LDR,
Rotary Pots, Lamp, Coupling T/fmr,
£6-65. PSU, £2-80. PCB (3½ in ×
10½in) Mk. 2, also holds 7 rotary or
slider pots, £1-75.

ULTRASONIC
TRANSMITTER-RECEIVER
(PE May 72). S/c's, Rs, Cs, Pot,
Relay, Dual PCB (2in x 5½in), £3-90.
T/ducers avail. to order, £6-30 per

ALSO
PCBs as published (while stocks last) (or: DIGITAL PSU (PE Aug. 72), 50p OSCILLOSCOPE P/A (PE Aug. 72), 33p SCORPIO (PE Nov./Dec. 71), 60p TRIFFID (PE Feb. 73), 60p

AND
DIGITRONIC (PW Mar. 73). Readout PCB (1½in × 3½in), 60p.

TAPE NOISE LIMITER (PE Feb. 72). S/c's, Rs, Cs, Pot, PCB (1½ in × 3in), £2.20. Reg. PSU and PCB (1½ in × 2½ in), £3.10.

VERSATILE LIGHT EFFECTS
UNIT
Single Channel Sound Controlled
Light with built-in variable strobe.
(PE June 72). S/c²s, Rs, Cs, Pots,
1/fmrs, Keyswitch, £8-85. PCB (3)in
x 73in) Mk. 2. also holds pots, Sw,
1/17 7/fmr, £1-50. SCRs—1A, 50p.

PHOTOPRINT PROCESS
CONTROL
(PE Jan./Feb. 72). For colour and B & W
—finds exposure, controls timing, stabs. mains voltage. S/c's, SCR, LDR, Rs, Colour Relay, Keyswitch, T/fmr, 47.50 FCB (3½ in ×5½in) also holds pots, Sw, relay, £1.20.

SOUND SYNTHESISER
(PE 1973)
Details of PCBs & Components in lists. SOME PRICES DOWN.

CAPACITORS **SEMICONDUCTORS ELECTROLYTIC** AC128 AC176 AD161 BC107 BC108 BC109 BC147 BC148 BC149 BC157 BC158 BC159 22p 13p 22p 20p 23p 22p 10p 10p 31p 40p

20p 27X531 20p 2X706 40p 2X706 40p 2X914 9p 2X1304 9p 2X1304 9p 2X1907 11p 2X3702 11p 2X3704 11p 2X3704 11p 2X3704 11p 1X3702 12p 1X5777 12p 1X9104 12p 1X4001 14p 1X 1.0/63 2-2/63 4-7/63 10/25 10/63 15/40 22/10 22/25 33/6-3 33/16 47/6-3 47/6-3 47/63 100/10 100/25 100/40 6p 6p 12p 16p 40p 10p 12p 20p 40p 45p 50p 110p 4p 6p 7p 8p 10p 17p 7p 6p 50p 36p

POLYESTER C280AE (μF/V) | TANTALUM BEAD (μF/V) 250V(µF) 0·1/35 0·22/25 0·47/35 1·0/35 1·5/35 0-01 0-015 0-022 0-033 0-047 0-068 0-1 0-15 0-22 0-33 0-47 0-68 1-0 1-5/35 2-2/35 4-7/35 10/16 10/25 15/6-3 22/16 47/6-3 47/16

POLYSTYRENE 63V 5% (ρF) 10 to 2200 3p; 3300 to 8200 4p; 10000 5p RESISTORS (E24 SERIES) \$\fomale 5\% C.F. \lip; \fomale \fomale 5\% CF. \lip; \fomale \fomale 5\% CF. \lip; \fomale \fomale 2\% M.O. \quad \text{Jp}

SWITCHES Makaswitch: Shaft asser assem 48p 33p bly Wafers Screens
pk of 5)
Spacers
pk of 10)
Rotary:
3P4W
Multipole (per 12p 30 p Lever: 4CL/4CL £1-30 4CL/4CN £1-30 Knobs 6p Slide: DPDT 24p

RELAYS
2PCO 12V £1
4PCO 12V £1-30
Relay base &
Clip 20p
Reed Relay
6-9V 85p base & 20p Relay 85p

NEW! RONDO (PE current series) PHASING UNIT (PF Sept. 73)

ENLARGER EXPOSURE METER AND THERMOMETER (PE SEPT. 73)

SEMICONDUCTOR TESTER (PE OCT. 73)

WIND AND RAIN EFFECTS (PE OCT. 73)

DETAILS IN LISTS

TRANSFORMERS (MAINS)

£2-47

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LIST

LIST

S.A.E. for free itemised list and with all enquiries please.

GENERAL
All PCBs Fibreglass, Drilled, Roller-Tinned. Layout and
Circuit Diagrams Free with each PCB. Unless stated "as
published", PCBs are designed by Phonosonics. Pots
are rotary unless stated as slider. All components are
brand new, top quality, guaranteed. Prices and deliveries
of individual components subject to stock availability.

Sub-min 12V-0-12V 50mA Min 0-6V 0-6V 6VA Min 0-12V 0-12V 6VA Min 0-20V 0-20V 6VA 17½W 1-6A 0-25-30 0-25-30V 1-5A

PHONOSONICS, DEPT. PEII, 25 KENTISH ROAD, BELVEDERE, KENT DAI7 5BW

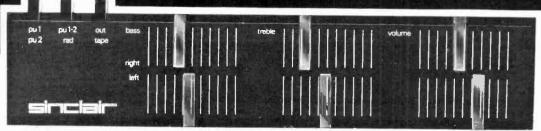
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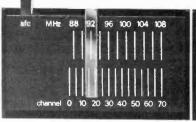
TXIOT

now... Project 80

--- exciting new thinking in modular hi-fi design



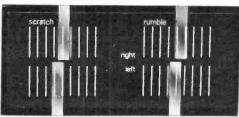
Stereo 80 pre-amplifier/control unit



Project 80 tuner



Stereo decoder



Project 80 Active Filter Unit (AFU)

the slimmest, most elegant hi-fi modules ever made



Living with hi-fi takes on new meaning now that Project 80 is here. These amazing new modules mark a brilliant technical advance all round; their size and presentation bring exciting new opportunities to install systems in ways hitherto only dreamed about but never before made practical. You can build a Project 80 system virtually anywhere and it is unbelievably simple to install and connect up. Everything that could possibly be wanted in a top quality do-it-yourself domestic hi-fi system will be found in Project 80 — compactness, elegantly ultra-modern styling, ease of fixing and operation, new control methods, and above all superb performance. New as well as popular established ideas on installation are featured on page four of this announcement to provide just a few examples of the system's fantastic versatility.



INTERNATIONAL AUDIO FAIR STAND 58

Project 80 new modules

Stereo 80 pre-amplifier and control unit

As with other Project 80 units, the Stereo 80 is mounted by means of two bolts fixed at the rear which pass through holes drilled in the wood or plastic on which modules are to be mounted. All the electronics are contained within the 3 deep front panel / Connecting leads are taken away similarly out of sight. Each channel in the Stereo 80 has its own independent tone and volume controls operated by sliders. This enexceptionally good environmental matching to be obtained Provision is made for magnetic and ceramic pick-ups, radio and tape in and out. A virtual earth input stage forms part of the up-dated circuitry of the Stereo 80 to ensure the finest possible quality from all signal sources. Generous overload margins are allowed on all inputs. Clear instructions with template are supplied

TECHNICAL SPECIFICATIONS

Size - 260 · 50 · 20mm (10 1 · 2 × 3 ins)

Finish – Black, with white markings Inputs – Mag P U 3mV RIAA corrected; Ceramic P U, 300 mV Radio 300 mV, Tape 30 mV

S/N ratio - 60db

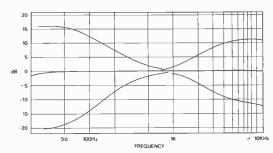
Frequency range - 20Hz to 15KHz · 1dB · 10Hz to 25KHz±3dB

Power requirements - 20 to 35 volts

Outputs - 100mV + AB monitoring for tape

Controls – Press button for tape, radio and P U selection Volume, Bass: 12dB to 14dB at 100Hz, Treble + 11dB to - 12dB at 10KHz

NEW For P.U., radio and tape Tape monitoring switch Two-hole fixing R.R.P. £11.95 + £1 19



Project 80 FM tuner smaller, more efficient

A truly remarkable tuner in every way - its unbelievably compact size its original circuitry - its dependable performance - all this in a boldly designed modern case measuring 85 - 50 · 20mm (3½ / 2 × ½ins). Greater adaptability (and possibly financial convenience) results from the tuner and stereo decoder section being made available separately

TECHNICAL SPECIFICATIONS

Size = $85 \cdot 50 \cdot 20$ mm (approx $3\frac{1}{2} \cdot 2 \cdot \frac{3}{4}$ ins) Tuning range = 87 to 108 MHz

Detector - I.C. balanced coincidence, for good A.M. rejection AFC - Switchable, with thermistor control to prevent from drift

One 26 transistor I.C.

Twin dual varicap tuning

Distortion – 0 3% at 1KHz for 75KHz deviation

Ceramic filter in I.F. section

Aerial impedance – 75 Ω or 240-300 Ω

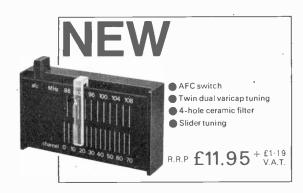
Sensitivity - 4 microvolts for 30dB quieting Power requirements - 12 to 45 volts

Project 80 stereo decoder

Making the Project 80 decoder separate from the F.M. tuner gives the constructor a wider choice of systems as well as saving money in cases where stereo reception may not be required. This unit gives a 40dB channel separation with an output of 150mV per channel. The gallium arsenide light emitting beacon automatically lights up to show when a stereo transmission is tuned in. Designed essentially as an integral part of Project 80 systems, this multiplex stereo demodulator may be used in many cases with existing single channel frequency modulated tuners to provide stereo reception

Size = 47 × 50 × 20mm (12 · 2 · 3ins)

One 19 transistor I.C.





new constructional techniques

. . . and again Sinclair leads the world

Micro-miniature power amp small enough to stand on a
10p. piece. Slimline pocket receiver smaller than a 20
cigarette pack

1963 Micro-6 receiver, smaller than a matchbox

1964 Pocket F.M. receiver; PWM amp.

Z.12 power amplifier module, PZ.3 power supply

Stereo 25 pre-amp/control unit

1967 Micromatic, Q.14 loudspeaker; the first Neoteric

1968 IC.10, the first ever integrated circuit for constructors' use

1969 Q.16 – improved version of Q.14 Systems 2000 and 3000: Project 60 launched

1970 IC.12: Project 605

Project 60 stereo FM tuner, Z.50; PZ.8

Improvements to Project 60 with Z.50 MK.2 and PZ.8 Mk.3 The Executive Calculator: Digital multi-meter

1973 Cambridge Calculator PROJECT 80 LAUNCHED

and next?

Project 80 active filter unit

This efficiently designed unit makes a highly desirable part of any worthwhile system where inputs may be from record, radio or tape. As with Stereo 80, separate controls are applied to each channel thereby making it easier to obtain ideal stereo balance in any kind of indoor environment

TECHNICAL SPECIFICATIONS

Size $-108 \times 50 \times 20$ mm $(4\frac{1}{4} \cdot 2 \cdot \frac{3}{4} | ns)$

Voltage gain - minus 0 2dB

Frequency response - 36Hz to 22KHz, controls minimum

Distortion - at 1 KHz - 0 03% using 30V supply

HF cut off (scratch) – 22KHz to 5·5KHz, 12dB/oct_slope L.F. cut off (rumble) – 28dB at 20Hz, 9dB/oct, slope

Z.40 & Z.60 power amplifiers totally short-circuit proof

Either of these entirely newpower amplifiers is intended for use in Project 80 installations although, of course, they are readily adaptable to an even wider range of applications. Both Z 40 and Z.60 incorporate builtin protection against shortcircuiting and risk of damage arising from mis-use is greatly reduced. Comprehensive instructions are supplied with each of the modules

Z.40 Technical Specifications Size - 55 × 80 × 20mm

 $(2\frac{1}{8} \times 3\frac{1}{8} \times \frac{3}{4} ins)$ 9 transistors Input sensitivity - 100mV Output – 15 watts RMS continuous into 8 Ω (35V), 30 watts music

power(into 4 Ω (30V) Frequency response - 10Hz-100KHz+1dB Signal to noise ratio - 64dR

Distortion – at 10 watts into 8 Ω

Power requirements - 12-35 volts

Z 60 Technical Specifications

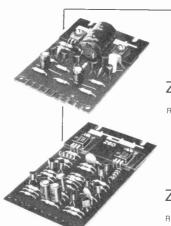
Size $-55 \times 98 \times 20$ mm $(2\frac{1}{8} \times 3\frac{3}{4} \times \frac{3}{4} ins)$ 12 transistors Input sensitivity - 100-250mV Output - 25 watts RMS into $8 \Omega (45V) 50$ watts music power

into 4 Ω (50V) Distortion - typically 0 03% Frequency response - 10Hz to more than 200KHz±1dB

Signal to noise ratio - better than 70dB Built-in protection against

transient overload and short

Load impedance – 4Ω min, max safe on open circuit



NEW

For scratch and

Transistorised

rumble control

active circuitry

R.R.P. £6.95

Z.40

R.R.P £5.45 + 0.54p

Z.60

RRP £6.95 + 0.69p

Sinclair power supply units PZ.8

the worlds most advanced unit in its class

Stabilised power supply unit. Reactainsed power supply unit. Re-entrant current limiting makes dam-age from overload or even direct shorting impossible, a principle never before inorporated in a commercially available constructor mod ule Normal working voltage (adjus table) 45V

R R.P. £7.98+0 79p V A T Without mains transformer PZ.5 30V unstabilised R.R.P. £4.98 + 0.49p V A T PZ.6 35V. stabilised R R.P £7.98+0.79p V A.T





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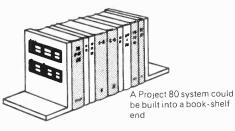
Practical Electronics November 1973

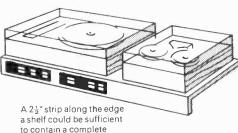
Recommended Project 80 applications

System	The Units to use	Units cost
Simple battery record player	2.40	£5.45 +54p V A.T.
Mains powered record player	Z.40, PZ.5	£10.43 +£1.04 V.A.T.
30W. RMS continuous sine wave stereo amp.	2 < Z.40s, Stereo 80; PZ.6	£30.83 +£3.08 V.A.T.
50W (8 Ω) RMS continuous sine wave de luxe stereo	2× Z.60s, Stereo 80; PZ.8	£33.83 +£3.38 V.A.T.
amplifier Indoor P.A.	Z.60, PZ.8	£14.93 +£1.49 V.A.T.
Car Radio	F.M. tuner, Z.40	£16,40 +£1,64 V.A.T.

From Sinclair the worlds most advanced hi-fi modules

Sinclair Project 80 the ultra-modern non-obtrusive hi-fi







The modules mount very easily onto a playing plinth



system

A novel application would be to build around the base of a lampshade



Project 80 could be easily mounted onto a loudspeaker cabinet



Two Sinclair Q.16 loudspeakers suitably positioned together with Project 80 could be mounted on to a false wall.

When you have seen for yourself how fantastically slim and cleverly designed these modules are, further ways will suggest themselves in which they can become a pleasing part of your particular domestic environment.

Please send post paid	
for which I enclose Cash/Cheque for £	including V.A.T.
Name	
Address	

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All prices inclusive of V.A.T.

 $P/P \ 12\frac{1}{2}p.$

Catalogue 10p.

(Saturday callers welcome)

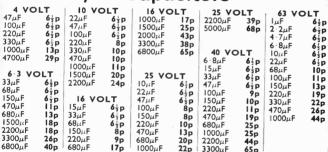
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SN7416 0.48 0.44 SN7446 £1 07 £1.04 SN7486 0.35 0.34	1
SN7417 0-48 0-44 SN7447 0-98 0-88 SN7480 #8.05 #8.70 Devices may	be
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Mullard Polyester's

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Carbon track 500 Ω to 2.2 meg Ω Log or Linear.
Single 13p. Dual gang (stereo) 44p.
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	10	0 plus
watt 5% carbon	1p	0·7p
watt 5% carbon	1p	0 · 7p
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range 10 ohms to 4.7		-
megohms.		
} watt ni/o 2%	3p	2 · 4p
range 10 ohns to 1		
megohms.		

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Small high quality type (linear only) All values 100—5 meg ohms, 0-1 watt 6p each 0-25 watt 7p each

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UP TO 25% OFF POPULAR SEW METERS

Type MR38P 1-21/32in S

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pe saa ioo x aomm.		
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Edgewise Meters

3-17/32in x 1-15/32in	x	2∄in	deep	
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12-24 volt

Volta	Amps	Volts	Amps	Part No.	Price
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le			1 5		

30 vol

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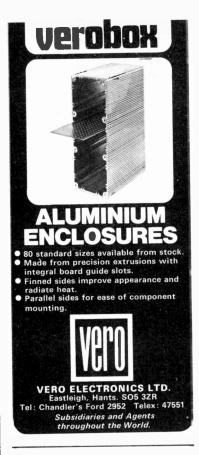
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2N2221A 2N2222		2N4919 2N4920	0-84 0-99	AF117 AF118	0.20 0.50	BC328 BC337	0.22 BFX84 0.19 BFX85	0-29	MJ2955 1.00 MJ3000 2.47	IN34A IN914	10p BA1	41 17p	BY237	121p OA79	7p
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2N2712 2N2713	0·12 0·17	2N5190 2N5191		AF178 AF179	0-55 0-65	BCY40 BCY42	0-81 BFY19 0-15 BFY20	0-35 0-50	MJE1092 1-98 MJE1102 1-57						
2N2714 2N2904	0·18 0·56	2N5192 2N5193	1.01	AF180	0-50	BCY43 BCY58	0.15 BFY29 0.21 BFY37	0-40 0-20	MJE2801 1-19 MJE2901 1-56	MINITRO	ELECTROI N 3015F 7-8	BEGMENT	POTEN Carbon:	TIOMETER	S
2N2904A 2N2905	0·70 0·71	2N5194 2N5195	1·10 1·46	AF186 AF200	0·40 0·35	BCY59 BCY70	0.22 BFY41 0.17 BFY43	0.43 0.65	MJE2955 1.65	INDICAT DRIVER	OR (16 PIN I	PIL) 22. 21-30	Log. or Lin	less switch, 1	7 jp.
2N2905A 2N2906	0-88	2N5245 2N5457	0-43 0-49	AF239 AF240	0·41 0·72	BCY71 BCY72	0.22 BFY50 0.18 BFY51	0-22 0-15	MM1613 0-48	SOCKETS	Light É	20p	Wire-wound	, with switch, id Pots (3W), 88	n. l
2N2906A 2N2907	0-70 0-79	2N5458 2N5459	0-46 0-49	AF279 AF280	0-54	BCY87 BCY88	8-47 BFY52	0-20	MM1711 0-49 MM1712 0-64	DIODE. (Red). 35	Made by TEX	AS INST.	Lin., 47p	ed Stereo Pots,	Log. or
2N2907A 2N2923	0.88	3N128 3N138	0.78 1.65	AFY42 AF211	0-74 0-55	BCY89 BCZ10	0.97 BFY56	0.15	MM2712 0-64 MPF102 0-25	MOTORO	LA MC SER	IES, RCA	BRECET	S (CARBO	
2N2924 2N2925	0-14 0-17	3N139 3N140	1·42 0·92	AL102 AL103	0-75	BCZ11 BD115	0.50 BFY76	0-41 0-22	MP8111 0-82 MP8112 0-40	CA 3000 8 PLESSEY	ERIES AND SL SERIES.	COSMOS, TRY US	0-I Watt	6p VERT	· / /
2N2926	- 1	3N141 3N142	0.81 0.58	ASY26 ASY27	0-80 0-86	BD116 BD121	0-50 BFY78	0-24	MP8113 0-47 MP8A05 0-25	FIRST FO	R LINEAR A	NDDIG1-		6p OI 71p HORIZO	R I
Green Yellow	0.12	3N143 3N152	0·75 0·92	A8Y28 A8Y29	0-28	BD123 BD124	0-82 BRY39	0.88 0.88	MP8A06 0-20 MP8A12 0-40	ENCAPSU	LATED FUI	L-WAVE			I
Orange 2N3053	0.82	3N 153 3N 154	0.81 0.84	A8Y50 A8Y55	0.20	BD130	0.57 BSX20	0·18 0·14	MP8A14 0-24 MP8A55 0-80		ER. 1-8A 100			DTENTIOME	TERS
2N 3054 2N 3055	0-86 0-75	3N159 3N187	1-17	BC107 BC108	0.16	BD131 BD132	0-40 B8X21 0-50 B8X26	0-20 0-49	MP8A56 0-28 MP8A65 0-25	2.5 watt	DUND RESIST		SINGLE GA	mm TRACK ANGED. LOG	or LIN
2N3390 2N3391	0.26	3N200 3N201	2-49	BC109	0-19	BD135 BD136	0-43 B8X27 0-49 B8X28	0.84 0.25	MP8A66 0-28 MP8U01 0-44	only), 7p 5 watt 5%). (up to 8-2kΩ	only), 9p.	lk to	IM. 30p each NGED, LOG	
2N3391A 2N3392	0.29	011401	1-00	BC113 BC114 BC115	0.12	BD137 BD138	0.68 B8X29 B8X60	0-47 0-54	MP8U05 0-48	10 watt 5	% (up to 261	Ω only),	lk to KNOBS, &	500k. 50p eacl	h
2N3393 2N3394	0.12	40050	0-78	BC115 BC116	0-15	BD139 BD140	0.88 B8X61	0-42	RETURN	S !!					I
2N3402 2N3403	0-18	40251	0.81	BC116A BC117	0.21	BDY10 BDY11	1.25 BSX77 1.50 ASX78	0·20 0·25	OF POST	Small		INI	CTD C	VTICO	
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