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Totally enclosed element in ceramic shaft. Fitted long-life iron -coated bit $3 / 32^{\prime}$
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Ditto tapped sec. 1.4 v.. $2,3,4,5,6.3$ v. $1 \frac{1}{4}$ amp. $\begin{array}{r}95 p \\ \hline\end{array}$ at $2 \mathrm{amp} . \mathrm{I}^{3}, 4.5,6,8,9,10,12,15,18,24$ and 30 v . £2-25 1 amp ., 6. 8, $10,12,16,18,20,24,30,36,40,48$. 60 . $£ 2.25$ 2 amp.., 6. 8. 10, 12, 16, 18, 20, 24. 30, 36, 40. 48, 60
5 amp., $6,8.10 .12,16,18.20,24.30,36.40 .48,60$ $3 \mathrm{amp} .5,8 \mathrm{and} 13 \mathrm{~V}$
$3 \mathrm{amp} .5 .5,8,10,13$ and $5-0-5 \mathrm{~V}$
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speaker from 8 to 15 ohms. Power consumption is less than I amp.
speaker from 8 to 15 ohms. Power consumption is less than I amp.
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| BC143 | 35p | $\begin{array}{ll}\text { BSX20 } & 15 p \\ \text { BSY27 } & 15 \mathrm{p}\end{array}$ | $\begin{array}{ll}\text { OC36 } & \text { 85p } \\ \text { OC42 } & 40 \mathrm{p}\end{array}$ | TIP41A 75p TIP42A 85p | 2N3053 2N3054 20p 50p | $\begin{array}{ll}40636 & 1.10 \\ 40430 & 1.00\end{array}$ |
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| ACl 32 | 12p | BC107 | 7p | 2N2222 | 13 p |
| ACI87 | 14p | BCIO8 | 7p | 2N2222A | 13p |
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 ces you ve ever wanted - for under -3 retail value. Up-to-the-minute slider controls for bass and treble. Jeparate foc Paice volume and balance conttols. Headphone socket on fiont panel. Push-button input controls - magnetic phono (high/low) tuner, sux, mono. monito
 Loudness push-button contral for peeffect sound at low output levels. Left and right push-bution on/off swithes for speakers. Noise filtering ${ }_{20-55} \mathrm{C}^{55} 5$ and tape monitoring farilities. Two auxiliary AC outtets. Frequency response $20-$ $20.000 \mathrm{~Hz} \pm 1 \mathrm{db}$ at fult power. 15 watts ims per channel. Walnut cabinet with satin aluminium trims. Inputs: phono 2.5 mV and 5 mV RIAA; tuner/aux 250 mV . Hum and noise: phono - 50db; iuner/aux - 65db. How's that for a soecification!

$O$
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 4 ${ }^{1}$ * wide.

## A

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This high quality apaek
IT has its own built-in
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giving you the idesal frequency responas for

hi-fi, natural or moad music listening. Is beauiful, hatry, oiled walnut cabinat
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 This is one of the lowesl pricad stero tunar amplifiers on the market. It covars the full range of both AM and FM broadcasi frequencies. And when you're switchad to FM. en indicator lights up whan a iterao signsi is recaived - that's the time to awitch to 'Sterso'I Tha SSA. 18 hat all the facilitise you'd ox. pect to lind on tunari costing twice as much - saparate volUma, bass, trable, balance and luning contrals. Solactor awitth for tapa, phono. AM. FM. sterso. Jack socket on front panal las storeo headphonas. Froquancy tange: FM 80.108 MHz. AM 535.1605 kHz . Fisquancy risponso: 50-10,000 $\mathrm{Hz} \pm 308$. ower oupur. will mole music powdinto two ohm


compact unit measuting only $5 \frac{1^{-}}{}{ }^{-}$wide. if
gh and of deep. itco ains own main
power supply, and has a ganged voluma control and apasata tratere controla for asth channal Sperilication: frequeney ras-
ponse $40-17000 \mathrm{~Hz}+3 \mathrm{~dB}$ : output 3.5 watls music power par ponse $40-17000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ : output 3.5 watls mutic power gar chamel into of ohms: input phono. 600 mV : signal-10-noisa atio benter than 45 dB

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## THE HY41



The HY41 supersedes the popular HY40 introduced by ILP last vear. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically $0.05 \%$ at 1 KHz into 8 ohms!! and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 etains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi -Fi but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY 41 and its various applications including a complete stereo amplifier.

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.
OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts
R.M.S. continuous.

LOAD IMPEDANCE: 4-16 ohms.
INPUT IMPEDANCE: 30 K ohms at 1 KHz .
VOLTAGE GAIN: 30 db at 1 KHz
TOTAL HARMONIC DISTORTION: less than $0.15 \%$ (typical $0.05 \%$ )
at 1 KHz .
FREQUENCY RESPONSE: $5 \mathrm{~Hz}-50 \mathrm{KHz} \pm 1 \mathrm{db}$.
SUPPLY VOLTAGE: $\pm 22.5 v o l t s ~ D C$.
SUPPLY CURRENT: $\overline{0} .8 \mathrm{amps}$ maximum.
PFICE: inc. comprehensive manual, P.C. board, five extra components and P. \& P.:MONO: $£ 4.90$

## UNILUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier to contain all feedback and equalization networks within an integrated pre-amplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use.

Inernally the HY5 provides equalization for almost every conceivable input, the desired function is achieved by use of a multi-way switch or by direct interconnection,

Two distinctive features of the HY5 are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from 16-25 Volts and a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo preamplifier.

Specifically and critically designed to meet exacting $\mathrm{H}_{i}$-Fi standards, the HY 5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

INPUTS
Magnetic Pick-up (within $\pm 1 \mathrm{db}$ RIAA curve) $2 \mathrm{mV} .47 \mathrm{~K} \Omega$
Tape Replay sexternal components to suit he*d). $4 \mathrm{mV} .47 \mathrm{~K} \Omega$
Microphone (flat) $10 \mathrm{mV}, 47 \mathrm{~K} \Omega$
Cerarric Pick-up (equalized and compen-
satable) $20-2000 \mathrm{mV}$. variable.
Tuner (flat) $250 \mathrm{mV} .100 \mathrm{~K} \Omega$
Auxileary 1250 mV . $47 \mathrm{~K} \Omega$
Auxiliary $22-20 \mathrm{mV}$. $100 \mathrm{~K} \Omega$

OUTPUTS
Main Pre-amp output 500 mV .
Direct tape output 120 mV .
ACTIVE TONE CONTROLS (Bexendall)
Treble $\pm 12 \mathrm{db}$
Bass +12 db .
INTEANAL STABILIZATION
Enables the HY5 to share an unregulated supply with the Power Amplifier.
SUPPLY VOLTAGE
16-25 volts
PRICE: MONO: $£ 3.60$ STEREO: $£ 7.20$

6 ma approx.


SUPPLY CURRENT
OVERLOAD CAPABILITY
better than 26 db on most sensitive input infinite on tuner and auxl.
OUTPUT NOISE VOLTAGE: 0.5 mV .

POWER SUPPLY PSU45
The versatile P.S.U. 45 is designed to supply your HY41's $+H Y 5$ 's in stereo or mono format.

Specification
Input: 200-240 Volts.
Output: $\pm 22.5$ Volts at 2 amps .
Overall Dimensions: L. 7"'; D. 3.8"; H. 3.1"
PRICE: $£ 4.50$ inc. P. \& P.

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# VOL. 8 <br> No. 12 <br> December 1972 

## FIFTY YEARS OF BROADCASTING

ANNIVERSARIES are always useful when they provide excuse to pause and look back over some really significant period. The pause is fully justified on the occasion of the BBC's 50th birthday. There is no question that these 50 years of broadcasting in the U.K. have had tremendous and lasting effect upon the individual and upon society as a whole. The history of broadcasting is also the history of the emergence of electronics, although the precise origins of both are indefinable or controversial, and go back certainly far beyond a half century.

In November 1922 the British Broadcasting Company was formed by a group of leading wireless manufacturers, at the behest of the G.P.O. This action followed years of agitation by various individuals and groups in this country. One of the most active pressure groups was the Wireless Society of London (later the Radio Society of Great Britain) representing the large number of amateur constructors and experimenters the new medium of wireless communication had already attracted. At the time, broadcasting was already in full swing in the United States where more than 300 registered stations were operating; in Europe regular broadcasts had been made from The Hague since 1920: while in this country irregular transmissions had been made throughout this period by the Marconi Company.

In 1926 the BBC was reconstituted as a public corporation by Royal Charter, fully committed to the public service concept by its indomitable chief executive J. Reith (later Lord Reith). High-minded principles and the uncompromising way in which they were applied produced a powerful, influential monopolistic organisation which has always been the centre of much controversy. The BBC during the past 50 years has been held up as an example of public service at its best, by some; while it has been condemned as an autocratic body out of touch with (or indifferent to) contemporary public opinion, by others. Political attitudes aside, all objective observers must acknowledge and pay respect to the integrity of purpose and high standards insisted upon and maintained consistently throughout these 50 years by the BBC , in both programme and engineering matters.
continued on page 1044

## THIS MONTH

## CONSTRUCTIONAL PROJECTS

RALLY DRIVER GAME ..... 1018
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P.E. ELECTRONIC PIANO-4 ..... 1038
P.E. DIGI-CAL-6 ..... 1050
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Our january issue will be published on
Friday, December 8

[^4]


Fig. 1. Simplified circuit showing principle of operation

## SIMULATION CIRCUIT

Fig. 1 shows a simplified schematic diagram of the whole unit. A 12 V positive supply is presented to VR1, the SPEED control, which is arranged in a simple voltage divider system. Since R1 is equal to VR1, the minimum voltage output is half the supply, i.e. 6 volts.

The resistor is used because if VRI was returned directly to earth, the input of the amplifier would also be earth and the output to the meter would be nil. Hence when starting off from rest the cars would achieve a speed of nothing; racing their engines without letting the clutch out, so to speak.

The next unit is a very interesting one that is really the heart of the system. It is a direct current amplifier with considerable negative feedback. Since it is direct current that is being amplified, capacitance coupling is out of the question and recourse must be made to direct coupling. The circuit used is a direct coupled d.c. amplifier which has a large input impedance, a very large gain, and considerable negative feedback, otherwise termed an operational amplifier.

## OPERATIONAL AMPLIFIER

Let us assume that the gain of the amplifier is 10.000 and that the voltage out is 10 volts. This would mean that the voltage at point " P " in Fig. 1 is $10 \div 10,000=0.001$ volt and considerable current is flowing back through $R_{\mathrm{f}}$ to maintain this very low figure-virtually at earth potential. It follows that increasing the value of $R_{\mathrm{f}}$ will reduce this cur-

Table 1: VALUES OF INPUT VOLTAGE TO ACHIEVE 10 V OUTPUT (where possible)

| $\boldsymbol{R}_{\text {in }}$ | $\boldsymbol{R}_{\mathrm{f}}=$ <br> $\mathbf{3 . 3 k} \Omega$ | $\boldsymbol{R}_{\mathrm{f}}=$ <br> $\mathbf{4 . 7 k} \Omega$ | $\boldsymbol{R}_{\mathrm{f}}=$ <br> $\mathbf{6 . 8 k} \Omega$ | $\boldsymbol{R}_{\mathrm{f}}=$ <br> $\mathbf{8 . 3 k} \Omega$ |
| :--- | :--- | :---: | :---: | :---: |
| $68 \mathrm{k} \Omega$ | 11.0 V | 8.5 V | 6.5 V | 5.75 V |
| $83 \mathrm{k} \Omega$ | $9.0 \mathrm{~V} \max 9.9 \mathrm{~V}$ | 7.5 V | 6.6 V |  |
| $97 \mathrm{k} \Omega$ | $5.4 \mathrm{~V} \max$ | 11.5 V | 8.6 V | 7.6 V |
| $118 \mathrm{k} \Omega$ | $1.0 \mathrm{~V} \max 6.8 \mathrm{~V} \max 10.6 \mathrm{~V}$ | 9.3 V |  |  |

rent, raise the voltage at point "P" and raise the output voltage.
$R_{\text {in }}$ has the opposite effect; raising this value will lower the voltage at point " $P$ " and hence the output voltage. The analogy to accelerator and brake is obvious; $R_{\mathrm{f}}$ becomes the accelerator and $R_{\text {in }}$ becomes the brake. See Table 1.

Of course $R_{\text {in }}$ and $R_{f}$ have upper and lower limits. .The upper limits are no problem, but by adjusting the lower limits we may readily obtain different responses to the variable resistors. These are set to the individual car characteristics. (See Table 2 and Fig. 2). It will be seen from Fig. 2 that there is a choice of fixed resistors of differing values. Which one is chosen depends upon which switch is pressed. Then there is a small fixed resistor common to all and finally VR2, which is the brake. The total resistance equals $R_{\mathrm{in} 1}$.



Fig. 2. Breaking is simulated by R 2 to R 6 switched in the brake position $R_{i n}$

Table 2: RESISTOR BIAS VALUES

| CAR (switch no.) | $\begin{aligned} & R_{\text {in }} \\ & \text { (brake) } \end{aligned}$ | $R_{\mathrm{f}}$ <br> (accelerated) | $\boldsymbol{R}_{\mathrm{c}}$ (max. speed) | $\boldsymbol{R}_{\mathrm{e}}$ (road holding) |
| :---: | :---: | :---: | :---: | :---: |
| 1 | 68k | 2.2 k | 4.7 k | 6.8k |
| 2 | 68k | 3.3k | NiI | 5.6k |
| 3 | 47k | 4.7k | 4.7k | 5.6k |
| 4 | 47k | 3.3k | 2.2k | 8.2k |
| 5 | 56k | 2.7k | Nil | Nil |
| VR | 50k | 5k | ' | - |

Fig. 3 shows the operational amplifier and its associated low output-impedance stage. The circuit is surprisingly simple; TR1 and TR2 form the amplifier. Since they are direct coupled as a super-alpha pair, their individual current gains are multiplied. Notice particularly from where the feedback is taken. At this point (TR2 emitter) the voltage rise is out of phase with the base of TR1 and hence negative feedback.

In the prototype, the car-selection switches were only two-pole (see components list), so another selection of fixed minimum values were not provided for the feedback. However, from experimental work that has been done on the prototype, and for those readers who might prefer to use three- or four-pole switches, values have been given in Table 2. Not much can be gained in this way, however, and those who wish to incorporate only two-pole switches can be happy in the knowledge that the system works well.
An emitter-follower TR3 provides a lowimpedance output that will not load the previous stages. The meter is arranged so that the resistance of the meter coil provides the emitter resistance.


Fig. 3. The basic operational amplifier with the practical feedback resistance $\mathrm{R}_{\mathrm{f}}$

Normally, the collector is returned to the positive rail directly, but the insertion of resistor $\mathrm{R}_{\mathrm{c}}$ in the collector lead will reduce the maximum output voltage. This is a useful facility for those who would like to limit the maximum speed of different cars; values are given in Table 2.

## ROAD-HOLDING

From the meter, the current goes to a push-button (S7 in Fig. 1) which is included for three reasons. Firstly, there will be many game moves in which a corner is not encountered; secondly, illumination of the lamp would give players a valuable clue as to the position of the meter; and finally, current consumption is very low at 6 mA , but jumps to over ten times that figure when the button is depressed and the lamp lights.
The corner selection switch S6 selects, in the first position, the battery on/off, while each of the other five positions selects a voltage-divider such as is shown as $\mathbf{R}_{\mathrm{d}}$ and $\mathbf{R}_{\mathrm{e}}$. All resistors in $\mathbf{R}_{\mathrm{d}}$ are connected together and fed to the Schmitt trigger, while those in $R_{e}$ are selected by the car press switches. Thus the actual voltage division that takes place depends upon the severity of the corner and the car selected. See Fig. 4.


Fig. 4. Switching for road holding ability

:
Fig. 5. Switch trigger and lamp driver

## SCHMITT TRIGGER

Transistors TR4 and TR5 in Fig. 5 form a Schmitt trigger that will trigger the output of TR5 when the incoming voltage rises to 1.5 volts. When TR5 turns on, TR6 switches off, hence nearly 6 V appears across the lamp, making it glow. The resistor in series with the lamp is a current limiter.

## CONSTRUCTION

The constructor should start by obtaining his bank of press switches. There are many types available and many suitable as far as the circuitry is concerned. The panel may then be cut and drilled to suit, following the dimensions given in Fig. 6. The hole for the meter may require adjustment as well as extra fixing holes.

The hole in the top right-hand corner is to view the lamp. This was just drilled and countersunk, but readers may prefer to use a coloured plastics lens.



Fig. 6. Front panel drilling details and label layout


Fig. 7. Details of the brackets and flap for the

## WIRING

All the switches, potentiometers and the meter can now be mounted on the panel and the wiring as shown in Fig. 9 undertaken. Seven wires lead off to the circuit board; leave these at least 4 in long. To save confusion, it is best if these wires are colour-coded. The battery clips are joined so as to connect the two 6 V batteries in series; the aegative wire is taken to the tag strip, and the positive to one central tag of S6.


Fig. 10. Layout of components on s.r.b.p. circuit board and underside wiring


Wiring of switches S1-S5

## CIRCUIT BOARD

It will be seen from the photographs that there is ample room inside the case and that a small circuit board will hold all the active components. The components are arranged in two distinct sections, sharing a common 12 volt supply line and a common earth. A suggested wiring layout is shown in Fig. 10.

Solder all the wires to appropriate points on the circuit board before putting it all into the case. Check all wiring against the circuits before connecting the batteries and testing. If for any reason the operational amplifier is not functioning as it should, adjustment of resistors in $R_{i \|}$ or $R_{f}$ or both may solve the difficulty.

When all is satisfactory it can be assembled in the case. The circuit board sits on top of the bottom front strip and on a similar strip at the back, while the batteries are held by similar wood strips glued to the case. The panel is fastened to the fillets by four $\frac{1}{2}$ in chrome-plated screws. Finally, a back of hardboard may be pinned into place, its edges chamfered and smoothed.

## COMPONENTS

## Resistors

| R1 | $10 \mathrm{k} \Omega$ | R11 | 47k $\Omega$ | R21 | $8.2 \mathrm{k} \Omega$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | $68 \mathrm{k} \Omega$ | R12 | 3.3k $\Omega$ | R22 | 10k S |
| R3 | $68 \mathrm{k} \Omega$ ~ | R13 | $33 \mathrm{k} \Omega$ | R23 | 100k |
| R4 | $47 \mathrm{k} \Omega$ | R14 | 22k $\Omega$ | R24 | 2.2k $\Omega$ |
| R5 | 47k $\Omega$ | R15 | 15k $\Omega$ | R25 | $1 \mathrm{k} \Omega$ |
| R6 | 56k $\Omega$ | R16 | $10 \mathrm{k} \Omega$ | R26 | 33k $\Omega$ |
| R7 | 33k $\Omega$ | R17 | 8.2k $\Omega$ | R27 | $47 \mathrm{k} \Omega$ |
| R8 | $6.8 \mathrm{k} \Omega$ | R18 | 6.8k $\Omega$ | R28 | $68 \Omega$ |
| R9 | $47 \mathrm{k} \Omega$ | R19 | 5.6k $\Omega$ | $\mathrm{R}_{\mathrm{c}}$ | ee tex |
| All $\pm 10 \% \frac{1}{4}$ watt carbon ${ }^{\text {a }}$ |  |  |  |  |  |
|  |  |  |  |  |  |

Potentiometers
VR1 $10 \mathrm{k} \Omega$ linear carbon
VR2 $50 \mathrm{k} \Omega$ log. carbon
VR3 $5 k \Omega$ linear carbon
Transistors
TR1, TR2 2N2926G
TR2, TR3, TR4, TR5 2N2926R or O
TR6 2N2196
Meter
ME1 Voltmeter 10V f.s.d. type MRA38 or similar

## Lamp

LP1 6V 0.06A m.e.s. with holder.

## Switches

S1 to S5 Push button 2 or 4 pole changeovers on common frame with reset, or separate toggle switches
S6 2-pole, 6-way rotary wafer switch
S7 Single-pole, push button on/off
Batteries
B1, B2 6 V each style PP1 (2 off)

## Miscellaneous

Battery connectors (2 pairs)
Aluminium panel $16 \mathrm{in} \times 4$ in
Plywood $\frac{1}{4}$ in for case and $\frac{1}{2}$ in square section strip
Component wiring boàrd or printed circuit board approx. $4.5 \mathrm{in} \times 4 \mathrm{in}$ or less
Glue, pins, knobs


Components mounted on rear of rally driver front panel

## GAME PREPARATION

A large sheet of card is obtained (the larger the better) and a road route marked out on it containing plenty of junctions, bends and corners of differing severity. Hazards, landmarks, buildings and other obstacles cant be incorporated to choice. It is worth marking lanes into spaces or car-lengths. Mark a starting/finishing line, and grade each corner and hazard.

## PLAY '

1
Toy cars are placed in each lane at the starting line. Players' may toss a coin for choice of lane. The last player to place his car has the privilege of moving off first. Each player specifies which type of car 'he is using and of course must stick to it throughout the game. A note is made of all choices. There is no limit to the number taking part, and any number of players may specify the same type of car.
In his turn, each player adjusts VR1 to his present speed. presses the switch for his type of car, then adjusts brake and/or throttle to his satisfaction. He lifts the flap and reads the voltmeter to the nearest whole division. Once he has lifted the flap he cannot readjust the controls until his next turn; he must lower the flap again afterwards. The player then moves his car the number of car lengths equal to the meter reading up to a maximum of ten lengths.

If during the course of this move he encounters a corner, he sets S 6 to the numbered grading of that corner. from 1 ' to 5 , with " 1 " being a slight bend and " 5 " the inside track of a hairpin. He then presses the button and if the lamp lights he spins off the track and misses the next turn. The turn after that he restarts from where he spun off.

## LANE DISCIPLINE

Only one car is allowed in a square at a time. Changing lanes is permissible, provided that in the lane into which he wishes to move or cross, there is no car nearer than four lengths behind; that is. there must be three clear lengths.
Cars wishing to overtake another may therefore be boxed in by a third and in these circumstances he must match his speed with the slower car in front. Failure to do so-that is, if his move would ordinarily take him beyond the car in front-means that he again must spin off the track and miss a turn.

Of course, players may use this technique deliberately to force a car behind off the road, and there will be many cases where a player over-compensates or grossly mishandles'the brake, bringing his car to , a halt in the path of another. Unless the one behind can overtake without hindering a third, or unless he too can halt, then he must spin off and miss a turn.

## CONCLUSION

Constructors preparing a track may have as many bends as they wish, but at least one good straight should be included-particularly if the top speed of some cars is limited in the manner described. Hazards such as chicanes and islands could also be introduced and constructors should bear in mind that the inside lane on a bend may have fewer carlengths in it, but rate a higher corner-severity.

Much of the success of the game and the enjoyment derived is in the road layout, and it will probably be worthwhile to collaborate on this with someone who has had real rally driving experience. $t$




## SOVIET SPACE PROBES

Some additional information has now become available from Russia which relates to the Mars-2 and Mars-3 space probes. Some of the work done by Mars-3 has already been reported here. The additional information was obtained with infra-red radiometers operating in the 8 to 40 micrometre band.

The system of measurement was such that the tracks of the probes went from the southern hemisphere to the northern hemisphere of Mars, going through the seasons and all parts of the day. The temperature of the surface of Mars was plotted through these periods and showed that the variation was from $-13^{\circ} \mathrm{C}$ to $-93^{\circ} \mathrm{C}$.

From this data a clue can be obtained as to the structure of the surface, and the climate can be assessed. The low temperatures at night seem to indicate that the surface cools quickly after sunset. If this is the case then it points to the heat conductivity of the surface as being low and consistent with dust or dry sand in a comparatively rarefied atmosphere.

Generally speaking, there was a difference between ihe dark areas or seas, and the light areas or continents. The seas were warmer by some $10^{\circ} \mathrm{C}$ and this is no doubt due to lower reflective power in these areas. This is true of one special area, a dark region known as Cerberus, which lies north of the Cimmerian sea. It is likely that the higher heat conductivity is due to rocks projecting from the surface.

In February when the probe Mars 3 was in the region of ihe northern ice cap the infra-red radiometer recorded a temperature of $-110^{\circ} \mathrm{C}$. Such a reading indicates that the polar cap consists mostly of solid carbon dioxide.

This cap remains all through the year. There is also indication of some water-ice. The southern ice cap disappears in summer.

## DRY MARS

The Soviet scientists have said that it is possible that the north polar cap contains more carbon dioxide and water than in the whole of the Martian atmosphere. The actual precipitation of water on Mars is about 0.02 of that of the earth's atmosphere. Mars. therefore, is much drier than earth based experiments and measurements indicated.

The dust storms on Mars still remain a mystery. Measurements have shown that the particles of dust are about one micrometre in size. These would settle very slowly, thus these dust clouds often endure for up to three months. One thing that has been noted with these clouds is that an increase in wavelength of observation can make them more transparent to the scanning photometer. Some particles are found to be as much as ten micrometres in size but these settle very quickly.

It is interesting to note that the two spacecraft from the Soviet Union have relied on astrophysical instruments for their observations whereas the United States have relied mostly on television. Such combination of observations should yield a much more positive accumulation of data for comparison.

So far the information from the various probes since 1965 has led to comparisons of earlier measurements. For example, in 1969 Mariner probes 6 and 7 recorded atomic hydrogen levels substantially higher than those recorded by Mars2 and Mars-3. The possible explanation is that either there is now a smaller amount of water vapour present. or that the temperature has fallen.

Finally, the presence of a shock wave due to the solar wind interacting with the upper levels of the Martian atmosphere, was detected.

## GALAXY THAT EXPANDS FASTER THAN LIGHT

Radio telescopes at California and Massachusetts. working as an interferometer with 100 million wavelengths as a base line, and enabling changes of 0.001 second of are to be detected. were focused on the source 3C 120, a quasar.

Astronomers D. B. Shaffer, M. H. Cohen. David Jauncey and K. Kellerman were examining the fine structure of this source and when they compared the present results with some made about eight months before, they observed considerable structural changes in the source. The movements observed, when
reduced to a measurement of velocity, indicates that the "blobs" of matter were moving apart at two to three times the speed of light.

There have been previous records in two quasar explosions which indicated speeds faster than light. However. the uniqueness of this new result lies in the fact that whereas the two previous events were quasars with a high red shift, about which some astronomers now have doubts, the results from $3 C$ 120 are from a source of quite low red shift. In this case there is no longer the criticism that the apparent high speed is because the distance from the earth is much less than supposed. The distance of 3 C 120 is not in question. The situation. has thrown the whole problem into a turmoil again.

## CRUST OF THE MOON

The artificial production of seismic events on the moon by crashing spent stages and lunar module ascent stages plus one authentic meteorite of about six feet in diameter, have now provided a great deal of data from the seismic observatories on the moon. Altogether there have been about six artificial impacts and these have enabled the reflections from thevarious layers of the moon to be measured.

There is a sharp increase in velocity from 0.1 km per second at the surface to 5 km per second at a depth of 10 kilometres. These figures are consistent with the structure of the moon soil and fragmented rock. As the depth increases the velocity would be expected to increase due to the compacting of the rocks. At a depth of about 25 km there is a sudden increase in velocity to 6.8 km per second and this continues down to a depth of 65 km . Down io 25 km the structure is consistent with basalt but from there on to the 65 km level there must be some other type of rock. At this level also there is another change in velocity which rises to 9 km per second. The geologists working on the projects maintain that this indicates a major structural boundary which may be the junction of the crust and the mantle.

The meteoritic impact revealed that the upper mantle velocity was similar to that of the earth with a velocity of 8.2 km per second. If this is so. the moon may have a mantle similar to earth.

The measurements also suggest that there is a core at a depth of about 600 km . This is the first direct evidence of a core. Having got so far, the last opportunity for direct observation comes in December with Apollo 17. At the moment it would seem that the crust of the moon is twice as thick as that of the earth.

$\square$

## by K. Lenton-Smith

ELECTRONIC MUSIC (as distinct from amplified conventional instruments) is often thought to be a relatively modern subject. In one sense it is, if Music Concrete is implied. In its real sense, the roots go back nearly 150 years, the recent pace having accelerated through technological advances.
C. G. Page described his music production methods in 1837 in the American Journal of Science in an article entitled "The Production of Galvanic Music'. In the following year, C. E. L. Delezenne published a paper describing how a contoured iron wheel, rotated in front of an electromagnet, could produce a musical signal.
Professor Thaddeus Cahill patented his Teleharmonium in 1897, consisting of 58 alternators. As amplification was then unheard of, each alternator produced a musical frequency which could be played directly into a telephone system: the sine wave outputs could be mixed to produce various tone colours-so pointing the way for those who were to lay down, some 40 years later, the basic principles behind many modern electronic musical instruments.

## MUSIC CONCRETE

The BBC Radiophonic Workshop and Daphne Oram are names that will be familiar to those interested in Music Concrete.

To produce this kind of music a tape recorder with 12 interchangeable capstans is required: with this, any sound may be recorded to produce a chromatic scale, provided that the source frequency is fairly steady. Re-recording this tape at different tape speeds on another machine will provide super- and sub-octaves of the sound source: careful cutting and splicing of these tapes can be done to produce a musical composition containing brilliant arpeggios and other unusual effects. A drum, cymbal (a very complex sound) or even a tap dripping into a basin are typical of the sound sources used in this work.

Rhythmic sounds are normally
produced by tape loops, perhaps using the basic source after its frequency has been taken down several octaves. After rhythm, melody, and special effects have been combined and reverberation added, the end result will either be fascinating or unpalatable-according to one's personal taste.
This type of work calls for several tape recorders and other special equipment, such as variable tone sources, re-iterators and reverberation devices. It is thus somewhat out of the amateur's field, except for the most rudimentary experimentation. The author admires the patience and skill of those prepared to make their music this way, but this type of work is not, of course, Strictly Instrumental.

## EARLY INSTRUMENTS

No real progress was made until amplification became a reality. Then, many ingenious inventions followed, most of which are now merely interesting curios.
The Trautonium, developed by Dr Friedrich Trautwein of the Radio Research section at the Berlin Academy of Music, was demonstrated in Berlin in 1930 and created considerable public interest.
The circuitry was fairly simplean amplified relaxation oscillator: later versions had some tonal variation and an improved fingerboard. It was the electronic equivalent of a one-string-fiddle, the chargedischarge characteristic of the neon tube being the same sawtooth as obtained by drawing a bow over a string.

Another early instrument was the Theremin, which employed the principle of heterodyning two r.f. oscillators to produce an audio beat-note. Frequency variation of one oscillator was achieved by hand capacitance, using a small antenna: the player thus had to "feel'" for his pitch, much like a string player. Though rapid passages were impossible, it was effective as a legato, solo instrument: vibrato and glissando were easy to accomplish.

Martenot produced a similar instrument, using a fingerboard to alter capacitance in the inductively coupled circuit.

## MONOPHONIC

It is easy to overlook the fact that a number of highly skilled musicians play monophonic instruments (brass and reeds, for example). Whereas the pianist or organist might be disinclined to consider a melody instrument, a monophonic device still has its uses.

The Hammond Solovox came on the scene about 30 years ago and was used by small bands as another solo voice. The fact that it soon had a number of imitators was proof of its success. The small, three-octave keyboard was designed to be fixed adjacent to the upper half of a piano keyboard, the master oscillator covering the entire 37 note range. Three divider stages followed and, with the master oscillator, their outputs were labelled, Soprano, Contralto, Tenor \& Bass. Any one-or combinationof these could be switched to a common busbar feeding a series of five tone filters. With a knee-operated swell and controls for attack and vibrato, the Solovox was a useful instrument in any small "combo'".

The monophonic instrument can only sound one note at a time (or its suboctaves), whereas the organ/ electric piano is fully polyphonic. In looking at earlier instruments, the development of polyphonic instruments has been ignored on this occasion: this is because the youngest member of the monophonic family-the Synthesiser-is important to many of today's musicians and constructors. The last few paragraphs may thus have helped, in a way, to trace its ancestry.

## THE SYNTHESISER

From the circuitry point of view, we have come a very long way from the Trautonium! The attraction of the modern synthesiser is, however, the same as it was over 40 years agoan entirely new sound. Multiple recording can produce existing, and often weird results.

The complexity of voltage controlled oscillators, envelope and noise generators, modulators, etc. means that the polyphonic synthesiser is still a twinkle in the eye of its commercial father. I.C. manufacturers are going through the throes of over-capacity: it may well be that they will turn their attention to simplifying the production of a polyphonic synthesiser at some future date.

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\because$ нау | 3 way | 4 way | ．${ }^{\text {way }}$ | fi whay | R way | 9 way | 10 way | 12 way |
| 1 pute | 40p | 40p | 40p | 40p | 40p | 40p | 40 D | 40p | 40p |
| $\because$ р川い号 | 40p | 40p | 40p | 40 p | 40p | 40p | 40p | 70p | 70 p |
| 3 pules | 40p | 40 p | 40p | 40p | 70 p | 70p | 700 | 95p | 950 |
| $\pm$ poter | 40p | 40 p | 10p | 70p | 70p | 70p | 70 p | \＆1．20 | \＄1．20 |
| $\overline{7}$ polew | 40p | 40 D | 70 p | 70p | 85 | 95p | 95p | \＄1．45 | £1．45 |
| 1）prolew | 40p | 70 p | 70p | 70p | 95p | 95 p | 85p | 21.70 | 81.70 |
| 7 pules | 70p | 70 p | 70p | 95p | 21.20 | \＄1．20 | \＄1．20 | \＄1．05 | \＄1．95 |
| ${ }^{*}$ pules | 70p | 70p | 70 p | 95p | \＆1． 20 | \＆1．20 | ¢1．20 | 22.20 | 1290 |
| $4{ }^{4}$ proler | 70p | 70 | ${ }^{965}$ | 95p | 21.45 | £1．45 | 21．45 | ¢2．45 | £2．45 |
| 10 peltax | 70p | 70 p | 95p | £1．20 | 21.45 | 11．45 | 21．45 | £2－70 | 22．70 |
| 11 pensm | 70p | 95p | 85p | 21.20 | \＄1．70 | £1．70 | 81.70 | e2．95 | 62．85 |
| 12 min | 70p | $95 p$ | 85p | £1－20 | £1．70 | £1．70 | 81.70 | 23.20 | $23 \cdot 20$ |

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## (3) Basic Operational Amplifiers

AN operational amplifier is essentially a high gain direct coupled amplifier which makes use of positive or negative feedback to control its overall characteristics. Originally the term operational amplifier denoted a circuit which performed a mathematical operation (such as, summation, differentiation, integration) in analogue computing. However, their applications are now extended to cover many fields; instrumentation and measurements are but two examples.

The required overall response of an operational amplifier is usually obtained by connecting feedback components externally from the output terminal back to the inputs. A differential input allows a considerable flexibility in the feedback configuration and also allows a reduction in d.c. drift because of the close matching of the input characteristics.

## BASIC FORM

The basic form of the operational amplifier feedback circuit, illustrated in Fig. 3.1, shows that the voltage at the virtual earth (or summing point) is very small and tends to zero for an infinitely large amplifier gain.

For this type of circuit where the output is fed back towards the input in such a way as to reduce it, the overall characteristics become almost independent of $A$ (the gain of the amplifier without feedback). The assumptions in the formula are that $e$ is zero, no current flows into the amplifier terminals, and that $A$ is much greater than $G$ (the overall gain with feedback).

Summation of currents at the virtual earth point must always be zero. The overall gain is given by:

$$
G=\frac{V_{0}}{V_{\mathrm{i}}}=-\frac{R_{2}}{R_{1}}
$$

The minus sign indicates a phase inversion.
Amplifiers of this type have a considerable isolation between input and output and many variations are

## clossany

OPERATIONAL AMPLIFIER A low frequency amplifier, often used to perform mathematical operations.
DIFFERENTIAL INPUT An input applied between two input terminals.
VIRTUAL EARTH Low impedance signal point. OFFSET ERRORS Errors produced by voltage or current imbalance in the input stages.
COMMON MODE INPUT RANGE The maximum signal which can be applied simultaneously to both inputs.
possible. The input impedance is equal to $R_{1}$. If $R_{1}$ equals $R_{2}$ the circuit gives unity gain where a sign change (inversion) is required.

## INVERTING AMPLIFIER AND ADDER

If a separate resistor is used for each input, as in Fig. 3.2, then current summing occurs at the virtual earth, with excellent isolation between inputs. For the first input the gain is

$$
G=-\frac{R_{\mathrm{FB}}}{R_{1}}
$$

Therefore
$V_{0}=-\left\{\left(\frac{R_{\mathrm{FB}}}{R_{1}}\right) V_{1}+\left(\frac{R_{\mathrm{F} 13}}{R_{2}}\right) V_{2}+\left(\frac{R_{\mathrm{FB}}}{R_{3}}\right) V_{3}+\ldots\right\}$


Fig. 3.1. Basic form for an operational amplifier


Fig. 3.2. Virtual earth addition amplifier


Fig. 3.3. Unity gain follower amplifier


Fig. 3.4. Non-inverting amplifier with gain


Fig. 3.5. Differential amplifier


TYPICAL DC CONDITIONS SHOVN (R2 = R3)
Fig. 3.6. AC coupled inverting amplifier


Fig.3.7. AC coupled non-inverting amplifier
The amplifier input bias current in Figs. 3.6 and 3.7 provides the voltage drop across the resistors R2 and R3. Since this current is low, then low leakage capacitors are essential

(a) DIRECT COUPLED

(b) AC COUPLED, DUAL SUPPLIES

(c) AC COUPLED, SINGLE SUPPLY

Fig. 3.8. Practical inverting amplifiers with ten times gain

## NON-INVERTING AMPLIFIER

For a non-inverting amplifier the input is connected to the non-inverting terminal of the i.c., while negative feedback is applied as shown in the diagrams. Feedback is now effectively in series with the input signal and care must be taken to make sure that the common mode inpul range of the amplifier is not exceeded.
A unity gain follower is shown in Fig. 3.3 where all the output voltage is fed back in series with the input.


Fig. 3.9. Practical non-inverting amplifiers with ten times gain

This gives a high input impedance, a low output impedance, and unity gain. The non-inverting amplifier with gain (Fig. 3.4) also has a reasonably high input impedance. The gain for this arrangement is

$$
G=\frac{R_{2}+R_{3}}{R_{2}}=1+\frac{R_{3}}{R_{2}}
$$

## DIFFERENTIAL AMPLIFIER

The amplifier in Fig. 3.5 uses both inputs to amplify the difference between two signals. The gain can be equal to, greater than, or less than unity.

$$
V_{0}=\frac{R_{F B}}{R_{1}}\left(V_{2}-V_{1}\right)
$$

## D.C. CONDITIONS

The amplifiers are designed to run from both positive and negative supplies with both input terminals and output terminal near earth potential (for a zero input signal) so that direct coupling can be used.

To maintain the correct d.c. conditions the amplifiers require a resistor from the output to the inverting input and a resistor from the non-inverting input to ground. Each input terminal should have the same d.c. source resistance to minimise the effects of input voltage and current offsets.

## A.C. AMPLIFIERS

Although the amplifiers are basically designed for direct coupled use they are equally suitable for a.c. applications. In these cases d.c. blocking capacitors are used in the signal paths. When electrolytics are used, care must be taken with the d.c. biasing conditions since many circuits maintain little or no bias across the capacitors.

Low leakage tantalum capacitors are preferred, or two high grade electrolytics may be used back to back. This problem can sometimes be avoided by using a single supply rail.

For the inverting ampiifier circuit of Fig. 3.6 the input impedance is equal to $R_{1}$ and the low frequency bandwidth is determined by R1 and Cl . For the noninverting amplifier in Fig. 3.7 the input impedance is approximately equal to $R_{3}$.

## PRACTICAL CIRCUITS

Some practical circuits for a.c. and d.c. use with a 741 C integrated circuit (or equivalent) are shown in Figs. 3.8 and 3.9. The connections given are for the TO5 version. The ceramic decoupling capacitors should be connected as close to the integrated circuit as possible. The capacitors in the a.c. coupled circuits with dual supplies only have a small bias across them and should be tantalum types.

Next month: Part 4 looks at one of the most useful i.c.s, the Schmitt trigger.

## BINDERS

[^5] W.C. 2 .


## By A. R. Cuff

Simple to build and capable of endless variations, this toy will provide hours of amusement for children and adults alike.

THIS project describes some simple circuits for producing sounds which somewhat resemble the cries of small animals. It should interest readers who are looking for something amusing. which is easily constructed, and can be used to intrigue their friends.
The "owl" described here is just one of a large number of possible variations of the basic design and makes an ideal present for children or the girl friend.

Because the idea is so versatile it can be made simple for beginners or rather more complex for advanced constructors; in fact this is a very good first project as there are no problems with tolerances or setting up, and only the cheapest components need be used.
The finished animal can be made extremely attractive if a little care is taken with the external detail.

## THE SIMPLE MODEL

The basic element in all the units is the astable multivibrator which is used in various combinations to produce a wide variety of animal and bird sounds.
In the simplest design as shown in Fig. 1 a single note astable is turned on and off by a mercury switch. The mercury switch (for those who have not encountered them before) is a sealed tube containing mercury which is free to move about. Two contacts sealed into the glass form a switch which is only closed when the mercury covers both. Thus when this switch is used in the model animal it can be arranged so that the animal only emits a noise when tilted

Also shown in Fig. 1 is a capacitively-coupled single-transistor amplifier to provide sufficient volume to operate a small 8 ohm speaker, and also to avoid excessive loading of the oscillator.


Fig. 1. With just a simple multivibrator and amplifier a toy animal can be built which emits a squeak each time it is lifted


Fig. 2. With two multivibrators in cascade, a more complex sound is produced somewhat resembling the miaow of a cat


Fig. 3. With three multivibrators in cascade the owl sound can be produced. Variation of the base resistors or capacitors in any or all of the multivibrators will produce different sounds

Thus the creature remains silent until picked up when it mysteriously squeaks!

## MULTIVIBRATOR DESIGN

In order to successfully design the astable of Fig. I, two important points must be kept in mind. Firstly, that the frequency of the oscillation is given by the expression

$$
\mathrm{f}(\mathrm{~Hz})=\frac{1}{0 \cdot 69\left(\mathrm{R}_{\mathrm{B} 1} \times \mathrm{C}_{\mathrm{B} 1}+\mathrm{R}_{\mathrm{B} 2} \times \mathrm{C}_{\mathrm{B} 2}\right)}
$$

so that the approximate frequency must be decided before construction is started ( 1 kHz gives a pleasing tone).
Secondly, to ensure oscillation the value of the base resistors must be small enough to provide sufficient base current to cause saturation of their respective transistors.

This can be written as $R_{\mathrm{H}} \leqslant h_{\mathrm{Fb}} \times \mathrm{R}_{\mathrm{C}}$ where $h_{\mathrm{FE}}$ is the current gain of the transistor.

However, since most constructors will be unable to measure the current gain it is safer to ensure that the base resistors are not more than 10 to 15 times the value of the collector resistors, especially if unmarked transistors are to be used. Thus if $\mathrm{R}_{\mathrm{C}}$ is 10 kilohms then $\mathrm{R}_{\mathrm{B}}$ should not exceed 150 kilohms.
For the sake of simplicity the base resistors are made equal in this circuit

## ADDING SOPHISTICATION

The next step to improve the sound effect is to add a second astable as shown in Fig. 2.
In this version the entire circuit of Fig: 1 is used as the collector load for another, slower astable. The frequency at which this circuit oscillates can be varied from about 1 Hz to approximately half the


Fig. 4. Layout of the components on the Veroboard panels. (Standard Veroboard was not used in the prototype. Miss alternate tracks if this layout is followed.)

## COMPONENTS

## Resistors

| R1 | 10k $\Omega$ |
| :---: | :---: |
| R2, R3 | $150 \mathrm{k} \Omega$ (2 off) |
| R4 | $10 \mathrm{k} \Omega$ |
| R5, R6 | $56 \mathrm{k} \Omega$ (2 off) |
| R7 | 10ks2 |
| R8, R9 | $120 \mathrm{k} \Omega$ (2 off) |
| R10 | $10 \mathrm{k} \Omega$ |
| R11 | $68 \mathrm{k} \Omega$ |
| R12 | $6.8 \mathrm{k} \Omega$ |
| R13 | $100 \Omega$ |
| All $\frac{1}{6}$ W | $\geq 10 \%$ carbon |
| Capacitors |  |
| C1 | $64 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |
| C2, C3 | $4 \mu \mathrm{~F} 10 \mathrm{~V} \mathrm{elec}^{+}$. (2 off) |
| C4, C5 | $1 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. (2 off) |
| C6, C7 | $0.003 \mu \mathrm{~F}$ (2 off) |
| C8 | $10 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |
| C9 | $64 \mu \mathrm{~F} 10 \mathrm{~V}$ elect. |

Transistors
TR1-7 BC108 or similar (7 off)

## Transformer

T1 Transistor audio output transformer (Eagle LT700 or similar)

## Switch

S1 Mercury switch

## Miscellaneous

LS1 $2 \frac{3}{4}$ in $8 \Omega$ speaker
B1 9 V battery type PP3
0.1 in matrix Veroboard, custard tin (or similar), felt or other suitable material for decoration

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| 600 | 12-24 | $4 \mathrm{c} / \mathrm{o}$ |
| 700 | 16-24 | 4M 2B |
| 700 | 16-24 | $4 \mathrm{c} / \mathrm{O}$ |
| 700 | 15-35 | $2 \mathrm{c} / 0 \mathrm{HD}$ |
| 700 | 16-24 | 6 M |
| 700 | 6-12 | $1 \mathrm{c} / 0 \mathrm{HD}$ |
| 700 | 20-30 | $6 \mathrm{c} / \mathrm{o}$ |
| 1,250 | 24-36 | $4 \mathrm{c} / \mathrm{O}$ |
| 2.500 | 36-45 | 6 M |
| 2,400 | 30-48 | $4 \mathrm{c} / \mathrm{O}$ |
| 9,000 | 40-70 | $2 \mathrm{c} / \mathrm{o}$ |
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73p
78p
63p
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Fig. 5 Construction of the owl. Felt or similar material can be used to cover the tin. The pattern shown is only a suggestion and variations can no doubt be devised by the constructor
frequency of the first to provide some very interesting effects. At low frequencies the "squeak" is switched on and off and sounds something like a cat miaowing, at higher frequencies a warbling sound is produced.

In the prototype the addition of capacitors CA and CB of 100 F each was found to reduce the sharpness with which the circuit switched, and also enhanced the sound considerably. Of course the value of the capacitors is open to experiment.


## HOOTING OWL CIRCUIT

By cascading astables as in Fig. 3, each time building on the existing circuit with a slower astable, very complex sounds can be made which are slow to repeat themselves exactly. (For more than three stages a higher battery voltage may be required.)

The circuit shown produces a sound somewhat like an owl, though other sounds can easily be made by changing the values of the capacitors in any or all of the astables.

A final point to remember when cascading stages is to ensure that each new stage can supply sufficient current to the existing stages. This means keeping the value of the base resistors as low as possible, in accordance with the desired frequency

## CONSTRUCTION

The scope for a suitable "carcass" is limitless, anything with sufficient space for the speaker and electronics will do. An ordinary $110 z$ size custard tin was found to be ideal for the owl.

First punch some holes in the metal end of the tin, and glue or solder the speaker in place (see Fig. 5). The circuitry is then built on Veroboard (Fig. 4) and soldered to the lid.

This method of construction allows sufficient room for the PP3 battery and the mercury switch as well as making access easy for battery replacement. The lid is just pushed home to seal the unit.

The outside of the tin can be disguised by using strips of coloured felt (see Fig. 5) and a very attractive finish can be obtained in a matter of an hour or so.

When completed the owl stands upright and is switched off for convenience. The mercury switch can be arranged to close at any desired angle to activate the device.

## ELECTRONIC PIANO

## Part 4

By A. J. Boothman B.Sc.



$\downarrow$AST month the construction of the basic pitch p.c.b. was described of which 13 are required. To test these and to compensate for possible variations in output from the logic circuits, it is desirable to make up a simple jig. Such a construction also facilitates choice of tuning capacitors for the Hartley oscillator.

## PITCH BOARD TEST JIG

The jig consists simply of a one foot square piece of wood to the edge of which is screwed two terminal blocks; one three-way and the other two-way.

The first block (3-way) connects to $\mathrm{V}_{\mathrm{s}}$, earth, and the pre-amp output on the pitch board under test, whilst the second block (2-way) anchors two floating leads at supply voltage. Response to keying action is checked by touching one of the floating leads against a key input pin on the pitchboard, whilst the other lead can be used to check the sustain action. See Fig. 4.1.

In order to ease the handling of the pitch boards during initial alignment, and to minimise the possibility of damage due to careless short circuiting of supply leads, etc. it is recommended that the p.c.b. frame is first built prior to any alignment.

## P.C.B. FRAME

Fig. 4.2 shows the method of construction of the frame. If the final mechanics of the piano are to fit together satisfactorily it is important that this is assembled with care to ensure squareness. To ease this problem it is suggested that screws alone are used at this stage and that the glueing process is left until later when the rest of the mechanics has been built.

## WIRING THE FRAME BLOCKS

The rear connector blocks, which have been cut into groups of three, carry the positive rail ( $\mathrm{V}_{\mathrm{s}}$ ), pitch output, and earth line. The pre-amplifier/ tremolo board is screwed onto the frame and panel, with the level tremolo depth and rate potentiometer spindles protruding from the frame side.

An eight-terminal connector is fixed to the bottom front bar at the right-hand side to provide the connections between the pre-amplifier board and other functions of the piano.

Red, blue and black wires for supply $\left(V_{s}\right)$, pitch output and earth return are connected to each rear terminal block. From the last block this wiring is made to terminate at the pre-amplifier/tremolo board.

The black and blue terminations are shown as a screened lead on the pre-amplifier board layout shown earlier in Fig. 2.7, but this is not necessary provided that the leads between the last connector block and pre-amplifier board are kept as short as possible. The red and black leads are left long at the opposite end of the module in order to provide connections to the power supply.

The eight-terminal block takes leads shown in Fig. 2.7 as follows. The red and orange, that will finally connect to S 3 ; the green, blue and red to S 2 ; and the screened lead to the main amplifier.


Fig. 4.1. Test set-up for individual pitch p.c.b.s

## TEST SET UP

With the pitch p.c.b. module completed as above the initial setting up is almost ready to commence.

Fig. 4.1 shows the interconnection of the various units to give a test and initial setting up arrangement for the pitch boards. In order to make the system operational certain components have to be added during the test phase as follows.

Since the foot and keyboard level controls have not yet been constructed they must be simulated by soldering a short jumper between the VR3 red and yellow wiring pins on the pre-amplifier board (Fig. 2.7), and inserting two resistors in the terminal strips corresponding to S 2 connections $\mathrm{A}-\mathrm{B}$ and $\mathrm{B}-\mathrm{C}$. The resistor values should be $1 \cdot 2 \mathrm{k} \Omega 2$ and $330 \Omega$ respectively.
As we are to test one pitch board at a time, the output load condition will be very different to the final situation when all boards are connected up so a $1 \mathrm{k} \Omega 2$ resistor is connected across the blue and black connections on the test jig terminal block.

From this point on the constructor could well be involved in a considerable amount of board removal and re-insertion, requiring repeated screwing
and unscrewing on the connector strips. It is strongly recommended that a good screwdriver be purchased for this purpose, and further that it should be the correct length for easy operation from above the pitch p.c.b. module.

## INITIAL SETTING UP

The purpose of the initial setting up procedure breaks down into the following:
(a) Determine whether any catastrophic shorts are present on any of the assembled pitch boards. Also make sure correct fuses are inserted in FSI and FS2 positions.
(b) Determine whether any mistakes have been made during board assembly, e.g. diodes connected the wrong way round
(c) Correct for any large variations in logic output if necessary.
(d) Correct for any large variations in individual note outputs due to transistor gain spreads if necessary
In order to approach the operation systematically, the key touch sensitivity is not involved at this stage



The thirteen pitch boards mounted in the wooden sub-frame
so that an acceptable balance is first achieved without the confusing element of touch variation.

## TEST PROCEDURE

A board is first wired up to the test jig, with an $0 \cdot 1 \mu \mathrm{~F}$ or $0 \cdot 2, \mathrm{~F}$ capacitor in position C 48 . On switching on the power, providing that everything is reasonably correct, a mixture of all the octaves of the note in question will be heard as a background noise.

Touching the test lead against any of the five key input terminals should cause the relevant note to trigger, and sustain in a fading manner for the period the test lead is held against the pin. Decay should be immediate on removal of the test lead, and it should be possible to simulate unkeyed sustain by connecting the second test lead to the sustain pin.

If, against the background noise, any particular note appears to be dominant, it may be desirable to reduce its level at this point. However some breakthrough will always occur, and the importance of this seems to diminish when the instrument is complete. It is therefore recommended that adjustments only be made if a note is extremely dominant. and the final decision as to the amount of time spent in this activity left until everything is built.

One point to remember here is that bass response is bound to be very low with the speakers mounted in this test arrangement with no cabinet resonance.

In order to adjust the breakthrough level on any particular note the value of R 74 (or $75 / 76 / 77 / 78$ ) is varied.

This is best carried out by soldering a second resistor across the $5.6 \mathrm{k} \Omega 2$ original to reduce its composite value, and thus reduce the signal to the relevant mixer. Convenient values to be tried are $56 \mathrm{k}!2,27 \mathrm{k} \Omega 2$ or $10 \mathrm{k} \Omega$.
Should any note require a higher output then a bridging resistor across $\mathrm{R} 84(85 / 86 / 87 / 88)$ can be used. When all the boards have been checked in this way they can be slotted into the pitch board module. and the $1 k!2$ resistor removed from the test jig. Tuning (and C48 selection) could be attempted at this point, but since this is done by ear it would be very difficult without the keyboard for easy interval reference.

## TUNING CAPACITOR

Capacitor C48 in the oscillator circuit is composed of a number of capacitors connected in parallel and whose values are chosen to give the required pitch at the central position of adjustment on the tuning
inductor. The range of capacitance required is approximately $0.1 \mu \mathrm{~F}$ to $0.4 \mu \mathrm{~F}$, and the selection process is carried out using the board test jig described earlier. The table gives suggested capacitor values with which to make the composite capacitor C48, but since winding the inductors by hand can result in quite a spread in inductance values, the capacitor values given should only be considered as a rough guide.

Table 1: Approximate capacity values for tuning pitch p.c.b.s

| Pitch | Frequency $\left(\mathrm{H}_{\mathrm{z}}\right)$ | Tuning <br> Capacitor (C48) | Added <br> Capacity $(\mu \mathrm{F})$ |
| :--- | :---: | :---: | :---: |
| F | 700 | 0.33 | - |
| F | 740 | 0.22 | 0.057 |
| G | 784 | 0.22 | 0.022 |
| Ab | 830 | 0.22 | - |
| A | 880 | 0.1 | 0.1 |
| B6 | 932 | 0.1 | 0.069 |
| B | 988 | 0.1 | 0.047 |
| C | 1047 | 0.1 | 0.033 |
| C | 1109 | 0.1 | 0.015 |
| D | 1174 | 0.1 | - |
| E6 | 1245 | 0.047 | 0.043 |
| E | 1319 | 0.047 | 0.032 |
| F | 1397 | 0.047 | 0.022 |

## PEDAL SOCKET BOARD

Figs. 4.3 and 4.4 show the circuits for the sustain and soft pedal sockets respectively. All components are mounted on the pedal socket p.c.b. The sustain voltage is dropped below $\mathrm{V}_{\mathrm{s}}$ by Zener diode D34. This gives a sharper fall in voltage when a key is released with the sustain pedal depressed, such that rapid playing of a single note in the sustained mode produces a noticeable attack each time the note is pressed. Capacitor C51 prevents switching noise.

The soft pedal is simply a potentiometer, but RI24 is connected on the p.c.b. as shown in order to present a similar load to the pre-amplifier when the soft pedal is disconnected. The piano can be played in this condition, and also with the sustain pedal unplugged. The jack socket is a stereo type in order to handle the three wires from the potentiometer.

In order to allow the pedals to be disconnected
from the instrument and packed away whilst being carried, they are connected to the piano by two screened leads and jack plugs. Mating sockets are mounted on a printed circuit board which is screwed to the base inside the piano, with suitable holes drilled in the base board to allow the plugs to pass through. See Fig. 4.5

When the pedal p.c.b. is later fixed to the base board, supply leads are taken from the pitch board frame, and the sustain output is connected to the nearest sustain point on the key frame assembly. The soft pedal socket is connected via screen leads to the pre-amplifier board. These are the green, yellow and red leads shown in Fig. 2.7 (part two), which should in fact be screened

## PEDAL SHELL DETAILS

The outer construction is common for both pedals, and consists of a shaped $\frac{1}{4}$ in plywood box, with an extra $\frac{1}{4}$ in strip at the front to act as a platform for the hinge. Fig. 4.6 shows the complete shell, including the $\frac{1}{8}$ in plywood flap.

The shell should be fixed together using a good grade of wood glue, and panel pins. The final assembly can be painted or covered in a fabric, with rubber foot and base pads.

## SOFT PEDAL DETAIL

There are many ways in which the action could be constructed, and the author would expect many builders to choose to use an alternative method to that described here

The important fundamental is that pressure should not be exerted on the potentiometer spindle against the body either in a rotational sense or at right

## SUSTAIN PEDAL



Fig. 4.3. Circuit diagram for the sustain pedal

## SOFT PEDAL



Fig. 4.4. Circuit diagram for the soft pedal

## COMPONENTS

```
Resistors
    R123 10k\Omega
    R124 2.2k\Omega
    R125 3.9k\Omega
    All resistors }\pm5%\frac{1}{4}\mathrm{ watt
```


## Capacitors

C51 $0.1 \mu \mathrm{~F} 125 \mathrm{~V}$

## Switches

S5 Single pole microswitch
S6 Single pole, 5 -way rotary switch

## Diodes

D34 12V 400 mW Zener diode
D35-D40 ZS170 (6 off)

Miscellaneous
JK3-Standard jack socket. JK4-Stereo jack socket. PL1-Standard jack plug. PL2-Stereo jack socket, knobs for keyboard control (40ff) (Bulgin)


Fig. 4.5. Component layout and printed circuit details for pedal socket p.c.b.


## MATERIALS LIST FOR FOOT PEDALS

```
Shell Construction (Two Required)
    5\frac{1}{2}
    5\frac{1}{2}
    3\frac{1}{2}
    3\frac{1}{2}
    5\frac{3}{4}}\textrm{in}\times4.4\textrm{in}\times\frac{1}{8}\mathrm{ in plywood (top flap)
    3in hinge
```

Soft Pedal Detail
$2 \frac{1}{2}$ in $\times \frac{1}{4}$ in dia, tension spring
3in $\times$ in in dia. wire
$3 \frac{1}{2}$ in $\times 2 \frac{1}{4}$ in $\times \frac{1}{16}$ in aluminium sheet
tin brass spindle coupling
Screws, spacers, and cable cleats as required
Sustain Pedal Detail
$1 \frac{1}{2}$ in $\times \frac{1}{2}$ in dia. compressing spring
3 in $\times \frac{1}{8}$ in dia. wire
$1 \frac{3}{4}$ in $\times \frac{1}{2}$ in $\times \frac{1}{16}$ in aluminium sheet (switch
actuator)
$1 \frac{1}{4}$ in $\times \frac{1}{2}$ in $\times \frac{1}{16}$ in aluminium sheet (retaining
bottom bracket for wire)
Screws, spacers, and cable cleats as required.
angles to the axis. The action should therefore only be built around the spindle, without clamping the body, but it is obviously necessary to provide a stop with which to prevent the body rotating.

Fig. 4.7 shows a possible solution. Lateral movement is prevented by positioning the support in such a position that the potentiometer body rests up against the shell side, whilst a small piece of wooden channel fixed to the base prevents rotating movement of the body. If brass bushes can be obtained, instead of relying on $\frac{1}{4}$ in holes in the aluminium, the action will be smoother.

The $\frac{1}{4}$ in spindle coupler is slipped over the spindle and secured with one screw, which also acts as a stop in the pedal off position. A further screw, with suitable spacer gives a lever on which to mount one of the cable cleats which acts as a bearing for the bent wire. The second cleat is fixed to the flap, and can be removed when access to the mechanics is required after the initial build.


Fig. 4.9. Touch control circuit


The two prototype foot pedals

## SUSTAIN PEDAL

The same comments apply concerning alternative sustain pedal mechanics as for the soft pedal control. In the prototype the author used a commercial footswitch, but details of a suggested assembly are given in Fig. 4.8.

A $\frac{1}{2}$ in diameter compression spring provides the power to return the pedal after operation, and the upward travel is limited by a wire shaped to be captive in the top bearing (cable cleat), and retained within a slotted hole in the lower bracket. Access to the internal mechanics is again achieved by disconnecting the top bearing from the flap. The spring is seated within $\frac{1}{2}$ in diameter holes drilled in packing pieces which are attached to the flap and to the base.

Depending on the microswitch obtained a number of alternative configurations might be used.

## TOUCH CONTROL

The touch control is mounted to the left of the keyboard, and gives preset selection of the maximum attack required at any particular time. The keyboard actuated touch sensitivity allows the player to reduce the attack below this maximum preset level by finger tip control.
The present variation in attack is achieved by controlling the voltage at the keyboard as seen in Fig. 4.9. Diodes D35-D40 give a range of attack voltages to suit the player. All diodes are mounted on the switch, which also has flying leads.

## LEVEL CONTROL

This acts as a stepped potentiometer which allows rapid changes to be made in the playing volume. The resistors are mounted on the switch S 2 , as shown in Fig. 2.4. Normal playing position is position 2.

## Nextmonth: Keyboard, keyswitch assembly and cabinet details will be given.

Two major technical developments, each to have profound effect upon the social habits of millions, are part of the exciting BBC story. They are television and recording

A late entrant into the field of sound broadcasting. Britain was to become first in the field of television. The all-electronic high-definition system developed by E.M.I. was adopted by the BBC when it commenced a regular television service in 1936, thus antedating the start of television in the U.S. by three years. So public service broadcasting stole at march on commercially-run broadcasting. This pioneering achievement demonstrated very forcibly the strength of a combination formed of a free enterprise industrial concern and a national institution devoted to public service, when working together with a definite objective in view

Recording has greatly influenced the pattern of broadcasting in recent times. In its earliest years the BBC made limited and catutious use of recorded sound (as distinct from commercial dises) although a Blattnerphone steel tape recording machine was installed at Savoy Hill as early as 1931. Apart from the imperfection of the techniques then available opposition to the use of recorded material was based on a strongly held view that actuality was the very essence of broadcasting. However romantic such a view may seem to blasé listeners of today, most listeners of the 20 's and 30 s agreed that the direct and immediate reception of events as they were happening was one of the unique enchantments of the new "wireless" entertainment.

Todays prodigious output of sound and television owes very much to the great advances made in the atudio and video recording techniques, notably due to contributions by commercial firms in Germany and U.S. All those hours of broadeasting we have become accustomed to can only be sustained with the aid of at stock of "convenience programmes" stored on disc or tape. With the insistent demand by the public for more broadcasting the "live" element is diminishing and thus the character of broadcasting is changing. There is a danger that broadcasting organisations will become merely carriers of other peoples" canned wares. (The automated sound broadcasting station with built-in magazines of audio tapes is already avaibable as a "packaged deal." How long before the television counterpart makes its debut?

This tendency will be accelerated with the introduction of commercial local radio next year, and with the possible granting of a second television channel to 1 BA . There is every reason therefore to ensure we preserve in its existing form our first and most experienced national broadcasting organisation with its unrivalled artistic production facilities that complement a justly renowned technical organisation. And champions of public service broadcasting do need to be vigilant. In anticipation of the expiry of the BBC's Charter and the Television Act in 1976. the pattern broadcasting will follow in the future is already undergoing careful examination.

## Talking Book Library for Handicapped

A national library has been established to help the in the normal manner. It is a library of talking books covering a wide range of subjects, all on purpose-made tape-cartridges containing up to 13 hours of reading.


The National Library of Talking Books is a registered charity and following a successful pilot scheme it already has a readership of 500 disabled people. It is hoped that Local Authorities and organisations for the handicapped will co-operate so that the talking books and playback machines may be made freely available to those in need. Further inquiries should be made to the National Library of Talking Books for the Handicapped. 49 Great Cumberiand Place. London WIH 7LH.

## Novel Educational Aid

An educational aid with a wide range of applications has just been introduced by Palca and is being distributed by Dixons Technical. Known as the Wireless Headphones System it makes use of a long loop of flex which surrounds an area within which communication is to take place. Anyone inside the loop and equipped with the special headphones, which are, in effect. self-contained receiver amplifiers, will hear the signal fed into the loop.

The loop acts as an indication loop aerial and its size can be extended to the requirements of the user. For instance it could surround a factory floor so that orders can be given in a noisy environment. or it could be used in a classroom with small groups of desks connected to different sources so that a number of lessons can carry on in the same room. Because the headphones are self-contained the wearer has complete freedom of movement allowing the system to be used for such applications as relaying information to footballers while the players are in motion.

The headsets cost only $£ 7.50$ each and the loop connector an extra $£ 2$ so making this a worthwhile and economic proposition.

Further inquiries should be addressed to Audio Visual Division. Dixons Technical Lid., 3 Soho Square. London
W1V 5DE.


The conventional way of checking ignition advance and engine speed is to use stroboscopic illumination of a timing mark on an engine part, such as the flywheel, the stroboscope being triggered by the lead to a sparking plug.

In BP1 273496 Snap-On Tools Corporation of Delaware, USA, suggests a more sophisticated approach.

Their system relies on a meter with a dial calibrated to show both engine and distributor advance in degrees and engine and distributor speed in r.p.m.

Instead of using a stroboscope to illuminate the timing mark, the mark is covered with a small strip of reflective tape and is illuminated by a sensing device housing a lamp and a photocell. During operation the lamp is eneraised continuously and the light from it picked up by the photocell only when reflected back from the reflective tape over the timing mark. This produces a series of voltage pulses and the patent describes circuitry for comparing this series if pulses with another series of pulses obtained from inductive coupling with a spark plua lead (ignition pulses).

Fig. 1 shows how the spark plua voltage pulses are fed, via amplifier transistor TR1, to a one-shot multivibrator (transistors TR2 and TR3). The pulses from the photo-

## BP 1273496



Fig. 1
cell XI are used to control the thyristor SCR1. A train of pulses triggered by the spark plua pulses are fed to the base of TR4, while another train of pulses triggered bv the photocell are fed to the base of TR5. The meter ME1 measures the time interval relative to engine period between successive pulses of the two pulse trains and converts this time interval into a direct read-out of dearees of enaine or distributor rotation.
As engine timing is automatically advanced with increasing enqine speed, the time interval between successive pulses becomes lonaer relative to rotation time and produces a higher reading in degrees on the meter. By reading the ianition advance in degrees for a series of enaine speeds and checking against those recommended by the manufacturer. the engine can be subjected to rigorous testing.

## WHEEL SPIN CORRECTION

Westinghouse Brake \& Signal Co. Ltd., have patented (BP 1277 474) some circuitry for use in testing devices for correcting spin or slide of a vehicle wheel.

These devices usually rely on deriving a rate of change of velocity signal depending upon the rate of change of the peripheral speed of the vehicle wheel. A modification circuit responds to a preselected value of this signal (indicative of the tendency of the wheel to spin or slide) and so modifies the braking forces.

A problem with such systems is that they are hard to check when stationary in a garage and the preselected signal values are hard to set. The Westinghouse invention attempts to tackle this problem.

The oscillator section of the patent is shown in Fig. 1. The unijunction transistor TR1 has $b_{1}$ base connected via R2 to the junction of R1 and Zener Diode D1. R1 and D1 are connected across a 30 V supply and $b_{2}$ base is connected via R3 and the primary winding of T1 to the supply negative line.

The transformer T1 is part of a test device coupled through the secondary winding to a generator which produces a signal of which the frequency is proportional to the peripheral velocity of the vehicle wheel.

The emitter of TR1 is connected to the junction of diode D2 and capacitor C1, the other side of D2 being connected via R4 to one
side of switch S1. Capacitor C2 bridges R4, D2, C1 and is itself bridged by a discharge resistor R5. An exactly similar arrangement of components is made between TR1 emitter and switch S2 with R7 acting as the discharge resistor.

As mentioned above, the secondary winding of T 1 is connected to the wheel generator of a spin or slide detection apparatus, the generator providing signals for the detection apparatus and the latter responding to a predetermined rate of change of peripheral wheel speed to initiate modification of braking.

As a service check a mechanic operates switches S1 and S2 to produce oscillations and inject them into the secondary winding.
Closing S1 causes oscillation by alternate charging of capacitor C1 via D2 up to a point at which the transistor conducts and discharges C1 via the emitter and $b_{2}$ base of TR1 into the primary winding of T1. The frequency initially depends on supply voltage, and in release of S1 changes at a rate dependent upon the rate of change of voltage applied to the emitter electrode of TR1 via C2. This is in turn dependent upon the discharge rate of C2 via R5. R5 is chosen to give a rate of change after release of S1 which corresponds to the minimum rate of change of speed of the wheel.

## BP 1277474



Closure of S 2 produces similar results but R7 is chosen such that the rate of change of frequency corresponds to a rate of wheel speed change to which the braking control is not required to respond.

Flipping switch S1 should trigger the modification circuit and flipping of S 2 should not trigger the modification circuit. If the mechanic finds this is not the case, then adjustment of the sensing and braking equipment is called for.


Areverse biased semiconductor diode has a high d.c. resistance. This is caused by the fact that the reverse voltage applied attracts current carriers (holes and electrons, in $p$ - and $n$-type semiconductors respectively) away from the junction, leaving a region near the junction depleted of holes and electrons, called the "depletion layer". The greater the applied reverse voltage, the wider is the depletion layer.

Since the portions that are not depleted of carriers have a low ohmic resistance, a reverse-biased pn junction behaves as a capacitor in which the electrodes (i.e. the non-depleted regions) are separated by a thickness of dielectric (i.e. the depletion layer). The thickness of the dielectric, and hence the capacitance. can be varied by varying the reverse voltage applied.

The same principles apply to all semiconductor pn junctions and, in particular, pn junctions in transistors. Thus reverse biased junctions in transistors may be used as variable capacitors by varying the applied voltage.

Silicon transistors are particularly suitable for such use. as currents across their junctions when reverse biased are very small, resulting in a capacitance effect of very low leakage. Bias voltages that can be applied to vary the capacitance, range from a small fraction of a volt to a maximum which is the break-down (Zener) voltage of the junction.
The a.c. voltage applied to the variable capacitor, however, must be a small fraction of the bias voltage, otherwise detuning and the generation of harmonics will result.

## TEST CIRCUIT

The circuit used for measuring the capacitance at different collector-base voltages is shown in Fig. 1. The inductance L1 is of known value. A reverse voltage is applied to the junction under test through the potentiometer VR1, and the resonant frequency of the circuit is determined by means of a grid-dip meter. A low impedance path for a.c. is provided by capacitor C 1 to the potentiometer wiper.

A number of transistors were tried. Plastic planar transistors appear to function best. Transistors encapsulated in TO5 or TO18 style metal cans had higher capacitances than plastic ones, although the ratio of maximum to minimum capacitance was much the same for both types.

Fig. 2 shows capacitances for applied voltages between 0.15 and 12 in transistors of three different types. There was some variation among transistors of the same type, but most transistors tested showed a ratio of maximum to minimum capacitance of about $5: 1$. The greatest change in capacitance occurred at low voltages.

Within the limitations mentioned these transistors functioned satisfactorally as tuning capacitors for frequencies up to 20 MHz or higher, and with reasonable care in the layout of circuits, the highest frequency tuned with a given inductance was generally a little more than twice the lowest.

## RECEIVER APPLICATION

The possible uses of voltage dependent capacitors are many, and include tuning, automatic frequency control, generation of harmonics, and others. As an illustration of the use of a transistor as a tuning capacitor, one was used in a simple reflex receiver for the medium wave band.


Fig. 1. (left). Circuit used to determine capacitance of a collector-base junction

Fig. 2. (right). The resultant voltage/capacitance characteristic of three sample transistors

The transistor used as a variable capacitor is TRI, in this case, type BCI08. The variable voltage applied to it is obtained from the potentiometer VR1; as most of the change in capacitance occurs at low voltages, VR1 should be a logarithmic type. Since the capacitance (and hence the tuned frequency) depends on the applied voltage, the power supply to the receiver should be stabilised by means of a Zener diode.

The inductor L1 used for tuning consists of 200 turns of 36 s.w.g. enamelled wire, close-wound on a ferrite rod $\frac{3}{8}$ in diameter and 3 in long. The secondary winding L2 consists of eight turns wound over the "earthy" end of L1, and wound in the opposite sense. The regeneration capacitor, C2 consists of two short lengths of thin p.v.c. insulated wire twisted together.

## TUNED CIRCUIT

Because of the small tuning capacitances and relatively high inductance used in tuning, C2 is connected to a tap on L1, 40 turns from the earthy end, rather than to the "live" end as usually done in receivers of this type.

Alternatively, C2 may consist of a short length of p.v.c. insulated wire; one end is connected to the collector of TR2, the other end is bent to fit snugly over L1 for one half turn. This half turn is moved along the length of L1 until optimum regeneration without oscillation is obtained over the whole tuning range, and is then fixed to L 1 in that position.

With reasonable care in the lay-out of components associated with TR1 and TR2 the receiver will tune over the greater part of the medium wave band. VR1 may be placed where convenient, preferably well away from LI to reduce detuning by hand capacitance while operating it. L3 should not be close to LI, and the axes of these inductances should be at right-angles to each other to avoid mutual coupling between them. Audio output is taken from C5 to high impedance headphones, or to an audio amplifier for loudspeaker operation.


Fig. 3. An example of the practical application of the junction capacitance of a transistor. In this simple reflex receiver the "varactor" effect is achieved by feeding a preset voltage to the collector-base junction of TR1

NEXT
MONTH....
A new area of interest, demonstrating how organic living systems and electronic systems can interact, is opened up with the publication next month of design and application data for a

## Biological Amplifier

for detecting very low frequency signals produced by the brain.
In conjunction with various output stages, this amplifier can be safely employed for specific non-medical purposes: as an alphaphone to induce relaxation, with aural feedback between the brain and ear; or as a cardiophone where control of the heart rate is achieved by means of a visual feedback loop.


Conventional ohmmeters have non-linear scales which, though providing large ohmic ranges, are difficult to interpolate and require special meter calibration. The I.C. linear ohmmeter gets round these difficulties using a cheap milliamp meter to provide three resistance ranges from $0-100 \mathrm{k} \Omega$.

## PRACTICAL



January issue on sale December 8
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# ELECTRONORAMA 

## AIDS TO MOTORING SAFETY

The problems of traffic control and collision avoidance on the world's roads has long been a problem that many countries have been seeking to find a satisfactory solution. Two companies, in the U.K. and America, have been developing separate systems, both based on the principles of radar. to help combat the problems of safety on the roads. The great advantage of using radar is that it is able to penetrate fog.

At the moment these systems are still in the experimental stages but both RCA and Mullard, the developers. hold great hope for the future.

## Radar "Cat's Eyes"

A traffic sensor has been devised at the Mullard Research Laboratories in Surrey, which could form part of a fully automatic system to control traffic, unaided by a human observer of traffic conditions.

One of the problems encountered with most radar traffic control systems has been that if more than one vehicle is in the beam it tends to confuse the equipment. But with the Mulard system the sensor device is very similar to the modern "cat's eyes" and is installed in the road in the centre of each lane. This way only the car passing directly over the sensor is detected.
The sensor device is a compact radar unit containing the transmitting aerial and receiving aerial and, as the cars passing overhead are detected, a second set of detectors activate circuits which measure the speeds of receding vehicles.

In a complete system, data derived from the radar "cat's eyes" installed in roads at strategic places would be routed to a central computer which would analyse the total traffic situation and control roadside warning lights and junction traffic lights accordingly.

## Anti-collision radar

The RCA experimental auto radar system is designed to avoid rear-end collisions on the roads and help drivers to maintain reasonably safe distances between the car and the vehicle in front.

A feature of the system is its ability to eliminate false "targets" and "clutter". It is not "blinded" by Tests being carried out on the RCA system. The radar units can be seen on the two cars. just above and below the licence plates



## Artist's impression of how one of the traffic sensors, being developed by Mullard, would be installed in the road surface.

the radar signals of vehicles passing in either direction in adjacent lanes. The radar beam is narrow enough to avoid unwanted reflection when a car moves out into another lane to pass a car ahead. Consequently. the car can pass without a warning being sounded.

The compact radar. mounted on the front of a car. transmits a continuous 9 GHz (X-band) vignal which strikes the passive harmonic reflector on the obstacle ahead. This reflector re-radiates the second harmonic ( 18 GHz ) of the incident signal. The radar receiver is designed to accept only those radar echoes that are double the frequency of the transmitted signal.

The reflector is approximately the size of a number plate and can be affixed to the rear of any vehicle or to any sign designating a traffic hazard. By using this special reflector all vehicles and traffic hazard signs, no matter what size. provide the same reflected signal strength. Thus a small car or a "one-way-do-not-enter" sign will be picked up by the radar system as clearly as a large container lorry. and their radar echoes will not be swamped by the echoes of larger vehicles.
Eventually, it is hoped, an operational radar of this type might be integrated into a car so that it would automatically release the throttle and apply the brakes to keep distances between vehicles above a safe minimum. Furthermore, if special small reflectors were incorporated on the sides of motorways and on collision hazards such as bridges, the radar could automatically apply the brakes if the vehicle is approaching the hazards at an unsafe speed.

The RCA system still requires testing and refinement, but scientists believe that an operational system can be mass produced within five years.
Showing how the RCA system works. Radar on car B transmits a signal ( $f$ ) which is reflected back at twice the frequency (2f) by a special reflector on the rear of the car A


Practical Electronics December 1972

# Featuras NiOP(-WNMY BAFETK and WEIGHING HAzAFDOUG MOVING LQADE 

## Automatic "Headway" Control

Ove example of automatic control in which brake and accelerator are operated as a result of obstacle detection, is being developed at the Lucas Group Research Centre near Birmingham.

Intended primarily for use on motorways where lane discipline is essential, this investigation has as its objective, the automatic maintenance of "headway" or the distance between moving vehicles at varying speeds.

The equipment uses a $35 \mathrm{GHz}(8 \mathrm{~mm})$ continuous wave source, frequency modulated to give the required information on range and speed. The use of this high frequency enables both the transmitting and receiving "horns" to be compatible with automotive styling requirements while providing the narrow beam width required. An electronic computer carried on the vehicle makes calculations based on the reflected radar signals and relative velocity of the vehicle with respect to the leading vehicle. Brakes and throttle are then automatically adjusted to maintain a consistently "safe" distance.
The Government Department of the Environment are supporting the project with financial aid and facilities for testing at the Transport and Road Research Laboratory


1. FMCW is frequency modulated continuous wave radar which measures "headway" and relative vehicle velocity
2. The relative velocity $k v$ caused by Doppler shift is such that it enhances the effect of headway $h$
3. Headway is compared with desired headway, calculated from time vehicle speed to produce an actuating signal
4. The actuating signal adjusts vehicle speed by means of brake and throttle actuators, so that headway spacing of one second is maintained

## Weighing-in-Motion Scheme for Heavy Loads

ONE of the problems encountered in heavy industry is the need to weigh large and sometimes hazardous loads on a continuously moving production line. Movement can influence erroneous measurements unless strict control of load vibrations is exercised. W. \& T. Avery lid the large weighing machine company have produced a "weighing-in-motion" system using a 27 ft load-cell weighbridge linked to electronic sampling. recording. and data transmission instrumentation.

Four 500.000 pound capacity load cells are equipped with free-motion units to isolate them from shear forces and transmit analogue weight signals to a static digitiser in the weigh office. An analogue to digital converter interfaces these signals to the digital display and teleprinter


High frequency vibrations are attenuated by the rigid design of the weighing structure and a filter in the instrumentation. A digital filter eliminates the effects of low frequency vibrations. By sampling the resultant output signals at various time intervals, the readout displays the loading in increments of 0.1 ton. If the load is moving over the weighbridge too fast, an "excess-speed" signal inhibits the system until reset.

This system has been installed at the British Steel Corporation's Port Talbot Works, where up to 600 tons of molten iron is weighed en route to the steelmaking plant. The load is hauled in huge ladles on a railway which passes over the weighbridge. <br> \title{
By R.W.COLES parts <br> \title{
By R.W.COLES parts <br> <br> aNSWER (A) AND $Z$ REGISTERS
} <br> <br> aNSWER (A) AND $Z$ REGISTERS
}

LASI month the construction of the ENTRY register was described together with the constant store and a fixed constant option. To follow this the logic and construction of the two registers which form the main part of the arithmetic section will be described. these being known as the answer (A) register and the $Z$ register


Fig. 6.1. Block diagram of the " $A$ " and " $Z$ " register sections of Digi-Cal

## "A" AND "Z'" REGISTERS

The a (aNSWER) and $z$ registers, together with the adder to be described next month, form the heart of the arithmetic unit of Digi-Cal. These registers are identical in capacity, being four (B.C.D.i digits "wide" and ten (decimal) digits "long". They differ in the type of input and output facilities which include both serial and parallel transfers.

Both are right-shift types. and are capable of shifting data at more than one million bits per second in the serial mode. when supplied with suitable clocking signals.

The A register has an associated display multiplexer which is used to send the answer data down the display bus under the control of the display strobes provided by the display board previously described in Part 3.

The multiplexer is identical in principle to the ENTRY register version, but is two (decimal) digits longer to allow display of a full eight digit answer. The most significant two digits are not displayed and are used to allow error detection and decimal point manipulation.

A block diagram of the A and z register sections of Digi-Cal is given in Fig. 6.1 and this should be studied in conjunction with the overall system block diagram (Fig. 1.3) from the July edition, which did not include the displaty multiplexer.

## REGISTER PARTITIONING

The division of the register circuitry to suit the constraints of a standard printed circuit system requires a good deal of thought to produce an arrangement which:
(a) uses a minimum of "wasted" i.c. positions, and thus the minimum number of separate boards;
(b) shares the number of input/output connections equally between boards so that one board does not demand more edge connector positions than are available on the standard board, and


Fig. 6.2. The partitioning of the " $Z$ " register is shown here, together with the inputs and outputs
(c) gives a layout arrangement which permits the interchange of identical halves of the circuit and thus cuts down the number of different types of board required.
With the ENTRY register last month, the circuit was divided vertically into three boards each carrying a two digit long register of B.C.D. width, plus a separate multiplexer board.

The $A$ and $z$ registers on the other hand are divided horizontally into two ten-digit long registers, each two digits wide, giving four boards in all. The A register multiplexer is also split in this way, and in this case is incorporated on the same boards as the register proper.

All this sounds a little complicated but is easier to understand with reference to Fig. 6.2 and Fig. 6.3.

## "A"' REGISTER

The A register is built on two DL109/44 cards, each of which houses eight TTL i.c.s, two resistors, and two decoupling capacitors.

Making a register ten digits long and four wide would require twenty SN7474 dual D-type flip-flops as used in the ENTRY register, and this number would take up a lot of room, as well as presenting a mammoth wiring-up problem.

Fortunately the TTL series of complex functions known as medium scale integration (M.S.I.) contains a number of ready-made shift register circuits, and it is one of these, the SN7496, which is employed in the a register circuit.

A diagram of the SN7496 is shown in Fig. 6.4, and as can be seen this i.c. consists of a complete five-bit serial shift register in a 16 pin D.I.L. package.

The register comes complete with gating to allow a five-bit parallel input transfer when the PRESET input is activated, and has a separate output from each stage to allow parallel output transfers.


Fig. 6.3. The partitioning of the " $A$ " register into the two halves known as the A1 and A2 registers is shown here

The clear input will set all the flip-flops to the logic " 0 " output state, and is used to erase the stored contents of the register.

By connecting two of these i.c.s in series a tendigit register is formed, and by using four of these ten-digit sections, the required a register capacity is achieved, using eight SN7496 i.c.s (see Fig. 5.5).

## OPERATION OF "A" REGISTER LOGIC

When an arithmetic programme is running, the data in the a register is shifted serially into the ADDER circuit, in synchronism with the data in the $z$ register which is either added to or subtracted from the a data.

The output of the ADDER forms the serial input of the A register, so that after a group of ten clock pulses, the a register contains the sum (or difference) of the original A and z register contents.

If the arithmetic operation required is multiplicaTION, then the contents of the a register have to be duplicated in the $z$ register before the series of additions take place, and to facilitate this the parallel outputs of the a register are used to provide a single 24 bit transfer.

At the end of a division, the answer is represented by the state of the counter, and before this can be displayed it has to be transferred to the a register. This is also carried out in a single 24 bit transfer, this time using the parallel PRESET inputs of the SN7496s.

In each of these transfers the initiation is provided by a control pulse from the programme.

## MULTIPLEXER

The multiplexer consists of a series of SN7401 open-collector gates, with their outputs "wired-or" to give the required four-bit data-bus. It is formed in the same way as the ENTRY register multiplexer.

The eight strobes or "character call-up" lines are produced in a sequence starting at the most significant end of the register, each one activating a group of four gates to send a single B.C.D. group to the display.
In the case of this multiplexer, the data-bus is produced in two, two-bit halves as a consequence of the partitioning arrangement mentioned earlier. This division of the multiplexer into two physically separate sections means that the display strobes have to be connected to both board Al and board A2.

## "Z" REGISTER

Like the a register, the $z$ register is housed on two DL109/44 plug-in cards. This register is identical in size to the a register, being ten by four in format, but because of its different role in the arithmetic operations of the calculator, different integrated circuits are employed.

A glance at Fig. 6.1 shows that the z register is required to accept parallel transfers from two sources, the a register and the entry register. It is not required to provide parallel outputs, only serial outputs being needed to feed the adder.
These different input/output requirements preclude the use of SN7496 shift register elements, since these devices can only accept a single parallel input transfer without extensive external gating. Luckily, it is still possible to enjoy the advantages of M.S.I. in this circuit because another device which suits our needs, the SN7494, is available.
The SN7494 is a four-stage serial shift register element with provision for parallel loading from two separate sources, under the control of internal gating.


Fig. 6.4. The internal logic of the SN7496 integrated circuit

A logic diagram of this device is shown in Fig. 6.6 where is can be seen that the two sets of parallel inputs, 1 A to 1 D and 2 A to 2 D are presented to the PRESET inputs of the flip-flops via four and-orinvert gates, the appropriate set of input data being selected by the presetl or preset 2 control line.

Note that output connections are not provided from the A, B, and C flip-flops, which as previously stated, is quite satisfactory for this application.

## REGISTER ASSEMBLY

Connecting two SN7494s in series gives a register eight digits long, which falls short of the requirement of ten digits. Therefore two more stages are added


Fig. 6.5. The logic contained on the A1 register board. The A2 board is identical to this


Fig. 6.7. The logic contained on the Z 1 register board is shown here. The Z 2 board is identical to this
in the form of an SN7474 dual D flip-flop. If these extra stages are placed at the "most-significant" end of the register, no extra input gating will be required since only the six "least-significant" stages are required to accept parallel input transfers. This is the approach adopted for the $z$ register and the resulting circuit is assembled from eight SN7494s and four SN7474s, giving the required total number of 40 flip -flops.

The circuit of one of the two identical $z$ register boards is shown in Fig. 6.7 where it can be seen that in addition to the shift register components, an SN7440 dual buffer gate is incorporated.
The connections to these gates are made external to the DL109 cards because these gates are used for different jobs on each of the two cards, and


Fig. 6.6. The internal logic of the SN7494 integrated circuit
internal connection would prevent interchanging these when required.

One of the four buffer gates is used to invert the CLEAR $z$ signal from the programme to make it compatible with the CLEAR inputs of the SN7474 devices, which operate on low inputs, as opposed to the SN7494 CLEAR lines which operate on high inputs.

The remaining gates are used to buffer the $z$ register presetl and preset2 lines, and in the overFLOW logic.

## "Z" REGISTER OPERATION

The most important operational aspect of the $z$ register is that it operates in a recirculating mode, which means that not only is its output fed to the ADDER, but also back to its own input, to give an endless loop.

This feature is important because it means that the register contents are not altered no matter how many bursts of ten clock-pulses are applied in the course of an arithmetic operation.
This facility is required in the multiply and DIvide programmes, where the contents of the $z$ register are continuously added to or subtracted from the contents of the a register until the result is produced.

In these programmes the contents of the register can be used up to one million times in a single computation. Of course, this is not to say that the register contents are completely invariant, for when a particular set of data is finished with the register will be cleared by a clear pulse from the programme, ready for new data to be entered in parallel from the ENTRY or a registers.

## A AND Z BOARDS



Fig. 6.8a. Layout of the i.c.s on the " A " register boards and function of the edge contacts both boards being identical


Fig. 6.8b. Layout of the i.c.s on the " 2 " register boards and function of the edge contacts, both boards being identical


Fig. 6.9. Disposition of the edge connectors as seen from underneath the main chassis plate

## CONSTRUCTION

The component layout arrangement for the $A$ and $z$ register boards is shown in Fig. 6.8.

These boards have a good deal of wiring to the edge contacts due to the large number of parallel input/output wires required, and some care is equired in laying this out so as not to produce very ong wiring runs or large wiring concentrations.
The registers are required to perform serial shifts at high speed which means that the interconnections associated with this aspect of the circuit, i.e. everything except the parallel input/output wiring, should be kept as short as possible to reduce stray capacitance effects. With this in mind it is advisable to wire up the serial interconnections and clock lines before the rest of the wiring (which is not critical) is added.

Wiring is normally carried out on the component side of the DL109/44 boards, but it was found with the prototype that wiring congestion can be alleviated on the $A 1$ and $A 2$ boards by connecting the parallel output wiring to the edge contacts on the underside of the board.

## EDGE CONNECTOR INTERCONNECTIONS

The edge connectors for these registers occupy the $\mathrm{Zl}, \mathrm{Z2}$. A1, and A2 positions on the chassis assembly (see Fig. 6.9), and are wired up in accordance with the wiring tables given. These tables (Fig. 6.10) are used by simply connecting the pin number required to the destination or destinations listed next to it in the table.

Each of these tables is complete, which means that a wire from the $A$ register to the $Z$ register will be listed in both the $A$ and the $Z$ register tables, so be sure to keep a record of wiring already carried out to prevent unnecessary duplication.

Again it is advisable to keep the wiring associated with serial transfers as direct as possible, and it is an idea to use a different coloured wire for these interconnections so that they may be easily identified later.

Where these serial data and clock lines exceed a few inches in length they should be routed as close to the chassis metal as possible to take advantage of its "ground-plane" properties. Interconnections to be treated with care are marked with an asterisk in Fig. 6.10.


Fig. 6.10. Interwiring details of the four edge connectors holding the "A" and " $Z$ '" register boards. The asterisks show connections which need special care


Fig. 6.11. Construction of a 44-way extension board which makes possible testing of boards in situ

## TESTING

Testing these boards in an operational way is difficult at this stage and until the clock board is constructed there is little that can be done except a check that the power supply lines are connected properly. For this reason it is essential that all wiring is double checked to remove wiring-up errors before proceeding.

## EXTENSION BOARD

When the calculator nears completion and tests are possible in an operational situation. it will some times be required to measure a logic level, or inspect an interconnection while the logic boards are still connected in the circuit. As things stand this would prove difficult because of the close packing of the DLI 09 cards in their respective edge connectors.

To overcome this problem it is possible to construct an extension board which, when interposed between a particular board and its edge connector, has the effect of raising the board into a position where all its components and associated wiring are readily accessible, while still electrically connected to the rest of the circuit.

The layout for such an extension is shown in Fig. 6.11 where it can be seen that the component parts are simply an edge connector and a Shirehall DL107 board, together with some wiring.

The DL 107 card is of the same physical size as a DL109, but instead of a printed wiring pattern to accept nine D.I.L. i.c.s, the DL 107 has a Veroboard type of matrix, orginally intended for wiring up discrete components.

The construction of the extension consists of simply soldering the 22 watys of one half of the edge connector to the DL107 strips which terminate in gold plated edge contacts, then using single core wire to connect the other 22 edge connector tags to the corresponding DL107 edge contacts on the reverse side of the board.

## COMPONETIS . . .

## A1, A2 BOARDS

Resistors
R67-R70 $1.2 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W} 10 \%$ carbon
Capacitors
C22, C24 $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. (2 off)
C23, C25 $0.047 \mu \mathrm{~F}$ (2 off)
Integrated Circuits
IC61-64, IC69-72 SN7496 (8 off)
IC65-68, IC73-76 SN7401 (8 off)
Printed Circuit boards and sockets
Type DL109/44 boards (Shirehall) (2 off)
Type DPK165 edge connectors (Shirehall) (2 off)
Z1, Z2 BOARDS

## Capacitors

C26, C28 $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. (2 off)
$\mathrm{C} 27, \mathrm{C} 29 \quad 0.047 \mu \mathrm{~F}$ (2 off)
Integrated Circuits
$\begin{array}{ll}\text { IC77, 78, 84, 85 } & \text { SN7474 (4 off) } \\ \text { 1C79-82, IC86-89 } & \text { SN7494 (8 off) } \\ \text { IC83, 1C90 } & \text { SN7440 (2 off) }\end{array}$
Printed Circuit Boards and Sockets
Type DL109/44 boards (Shirehall) (2 off)
Type DPK165 edge connectors (Shirehall) (2 off)
(In each case components are divided equally among the two boards which make up the A and Z registers)

The physical strength of the assembly can be improved by using a fillet of Araldite to join the edge connector and the DL107.

This simple servicing aid can come in very handy when testing any of the twelve Digi-Cal logic boards.

## Next month : Clock Board.



## COLD SHIVERS

Cold shivers ran up and down quite a lot of spines for a few brief moments when Dave Griffin, Motorola's director of marketing, Northern Europe, suggested there might be 'another slump in the semiconductor market in 1975. There was a sharp turndown in business in 1966/67 and the 1970/ 71 disaster is still painfully fresh in our memories. Is there a fouryear cycle of slumps? Could be, says Griffin. Of course nobody really knows but Griffin's remarks got everybody. crossing their tingers.

Griffin was speaking at a one-day seminar organised by component distributor Celdis Lid. A hundred delegates turned up to hear the speakers in the Sculpture Gallery at Woburn Abbey. Their wives and girl friends were invited too, but they were hived off during the business sessions and enioyed themselves watching wildlife in the Game Park and admiring the treasures in the Abbey itself.

But the sessions were by no means all gloom. Griffin came clean on Motorola forecasts which, among a great pile of statistics gave a U.K. semiconductor market worth $£ 80$ million by 1974. Big growth area was confirmed to be MOS devices which should reach £10 million worth of business in the same year. Big surprise is the market for germanium power devices which was expected to drop off very severely but which still has £4 million worth of business a year, even today.

This year, said Griffin, the U.K. market will have bought 40 million integrated circuits and 600 million discrete devices. Average selling prices have dropped since 1968 . Then, the average price for a discrete device was 25 cents and for
an integrated circuit 1.8 dollars. Today the averages are 20 cents and 1.2 dollars, respectively (dollar values are used universally in this type of comparison in the semiconductor industry).

## TIME AND MOTION

The automotive industry as a major consumer of semizonductors will not really take off until 1975. But electronic watches are already starting to be a good growth area. Motorola has an electronic kit which is sold to watch manufacturers. It consists of a quartz crystal, a CMOS oscillator, divider and buffer circuit equivalent to 312 transistors in a 6-lead ceramic package, and a microminiature stepping motor which drives the hands. Total power consumption is a mere 5 microamps from a 1.5 V silver oxide battery which lasts for over a year.

Motorola has greatly improved the yield of the CMOS package for watches by the use of ion implantation in which dopant material is "fired" at the target at enormous velocities but in a manner which can be deliberately controlled.

Another of the speakers, Kim Morrell of Mullard, forecast expansion of the U.K. electronics industry to $£ 1,800$ million by 1980 of which some $£ 260$ militon would be in components of all types. My own belief is that his figure is on the low side and we should do much better than that.

The Vice President of RCA Solid State Europe, Dr Joe Donahue, spoke on his company's solid commitment to European operations and to broadening the market base. The decision to set up in Europe was taken only three vears aqo. Now, says Donahue, RCA Solid State has almost 1,000 emplovees in Europe, well over 500 of them at the new semiconductor plant at Liege.

## TOGETHERNESS

On the theme of togetherness, Donahue reminded his audience that RCA grew out of the old British-owned American Marconi Company. That was way back in 1919 when RCA was formed at the request of the American goverhment to acquire the Marconi interests. It was the first time I ever heard RCA even admitting British ancestry publicly, let alone boasting about it!

To round up an interesting day, Peter Turner of Hewlett-Packard spoke of the dramatic arowth of the optoelectronics market. Apart from obvious uses such as displays, Turner stressed that a huge arowth area lay in electrically isolated couplers. H-P has a new fast one
on the stocks promising a 20 MHz bit rate with 2.5 kV isolation, all in a small i.c. package. Overall optoelectronics sales in the U.K. could soar to $£ 150$ million per year by 1976 stated Turner.

So, on the whole, the news was good, especially for distributors who are currently handling some 20 per cent of all professional component supplies and are working hard to handle more.

## CHATAWAY AT H-P

It was a great day for Dennis Taylor, managing director of Hew-lett-Packard Lid., when the Minister for Industrial Development, Mr Christopher Chataway, officially opened the $100,000 \mathrm{sq}$ ft extension to their South Queensferry plant. It more than doubles the area of the original unit built in 1966 and total employment at the plant has now risen to over 600.

Their total business in the U.K. is some $£ 9$ million and of total U.K. production 66 per cent is exported. Most famous of their instruments designed and built in the U.K. is the Microwave Link Analyser. To date over $£ 4 \frac{1}{2}$ mitlion worth have left the factory and 85 per cent went for export.

On the future, Taylor announced that they have just bought 33 acres of ground so we can expect further expansion although, chuckled Taylor, it will take a little time to expand into it. Hewlett-Packard has an $R$ and $D$ group of 50 people at South Queensferry and at least two completely new UKdesigned instruments will be released by the end of the year.

## RESETTLEMENT

An ex-editor in electronic publishing, $T$. Jeffrey Burton, and now a press relations consultant, has recently run a one-day seminar for RAF electronic technicians nearing the end of their service.

The idea was to give them a teach-in on the structure of the electronics industry and the lob opportunities within it. Speakers include representatives of the electronics press, of the Society of Radio and Electronics Technicians (SERT), the Post Office and companies of the status of IBM and Exchange Telegraph.

Although the scientific and $R$ and D ends of the business are not recruiting heavily at the moment, it seems that good technicians are hard to come by in the industry and are eagerly sought. The resettlement briefing, held at RAF Newton, near Nottingham, should help the boys in blue to ease themselve's back into civvy street and do a good turn to the electronics industry as well.


A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.
This is YOUR page and any idea published will be awarded payment according to its merits.

## VISIBLE SPEED READOUT FOR A PROJECTOR

Many home movie projectors rely for speed control on a rheostat which gives a film transport speed continuously variable from around 12 f.p.s. to around 25 f.p.s. This kind of control can have advantages over fixed (i.e. switched) controls, because many of the old silent films were shot at film speeds somewhere between 16 and 24 frames per second. Moreover, not all home movie cameras run at the same speed.

But anyone projecting films with a rheostat control will know that the snag with these controls is their unreliability as a positive guide to actual film speed. The same rheostat setting cannot be relied on to produce exactly the same speed on two occasions.

A few years ago I used a car tachometer or rev. counter to provide a direct visible readout of film speed.

Tachometers function by sensing the voltage pulses proportional to projector shaft speed; it is only necessary to produce low voltage pulses from the projector itself and'feed them to an inductive load across which the tachometer is connected.

Experience shows that the most reliable source of pulses is from a make-and-break contact associated with the projector main drive shaft, but of course any positively driven shaft will do. All that is necessary is to attach a small spring metal strip (such as an old relay contact strip) to an insulating block which is secured (e.g. glued) to the projector chassis so that the spring strip presses against the shaft (Fig. 1).

A blob (or blobs-see later) of Araldite or other hard insulating material is painted on the shaft so that, as it revolves, the spring strip will alternately make with the metal shaft and break by contacting an insulating blob. By connecting leads between the spring strip and projector chassis (which will of course be electrically connected to the shaft) a make-and-break connection is easily achieved which can be used to produce low voltage pulses when in circuit with a low voltage source.

Tachometers or rev. counters come with two leads, one of which is red or marked "positive" for connection to the positive side of the supply.

The tachometer dial reading will not of course necessarily give a true r.p.m. reading for the projector shaft, but this is not important. By timing a film loop through the projector and marking the dial accordingly, points of reference for actual speeds of 16,18 and 24 frames per second can be marked on the dial and the projector brought reliably to these speeds at any time in the future by operation of the rheostat.

By virtue of this arbitrary scaling, there is wide latitude of components, virtually any inductance or low voltage d.c. source that produces a healthy dial reading being acceptable. For guidance, 9 or 12 volts smoothed d.c. from a battery eliminator should be fed through the make-and-break contacts to the inductive load, for which the primary of a 5:1 transformer has proved ideal. To avoid radio or television interference, a 0.1 or $0.01 \mu \mathrm{~F}$ capacitor should be connected across the contacts.

A. Hope,<br>London<br>N.W.3.



Fig. 1. Arrangement of the spring contact on the projector drive shaft. The insulating spots on the shaft can be any hard-setting glue

Fig. 2. Typical set-up for the visible speed control readout


RST
VALVE MAIL ORDER CO． 16a WELLFIELD ROAD，LONDON SWI6 2BS SPECIAL EXPRESS MAIL ORDER SERVICE
Express postage piper cransistor，over cen
INTEGRATED CIRCUITS $5 p+1 p$ each added

\begin{tabular}{|c|c|c|c|c|c|c|c|}
\hline 1N21 \& $$
\begin{aligned}
& 80 \\
& 0.17
\end{aligned}
$$ \& AドZ1． $\begin{gathered}2.9 \\ 1.00\end{gathered}$ \&  \& \& $$
\ell_{0}
$$ \& \& 39 <br>
\hline 1 N 23 \& 0.20 \& ASY96 0.25 \& BY＇711 0 \& OAZ292 \& \& 28170 \& 0.10 <br>
\hline 1N85 \& 0.88 \& $\begin{array}{ll}\text { ASY } 27 & 0.82\end{array}$ \& BYZ1．0 0.80 \& OAZ223 \& 0.45
0.45 \& 28271 \& 0.18 <br>
\hline 1N253 \& 0.50 \& AsY28 0.25 \& B BLI 0－80 \& 0 －${ }^{\text {O22 } 24}$ \& 0.45 \& 2T21 \& 0.25 <br>
\hline 1N266 \& 0.50 \& ASY29 0.80 \& BYZ

0 \& OAZ241 \& 0.22 \& 2T43 \& 0.85 <br>
\hline 1N645 \& 0.25 \& ASY36 0.85 \& BYZIS 1.00 \& OAZ242 \& 0.23 \& ZTX 107 \& 0.18 <br>
\hline 1N726． \& 0.20 \& HAY50 0．17 \& BYZ10 0.62 \& OAZ24 \& 0.22 \& ZTX108 \& 0.18 <br>
\hline $1 \mathrm{NOL4}$ \& 0.07 \& ASY5 0.40 \& ВY78803 ${ }^{\text {¢ }}$ \& OAZ24 ${ }^{\text {a }}$ \& 0.28 \& 7TX 300 \& 0.12 <br>
\hline 1N400： \& 0.20 \& ${ }^{\text {Asy }} 530.20$ \& 0.15 \& 0 az290 \& 0.38 \& ZTX304 \& 0.25 <br>

\hline $$
\begin{aligned}
& 18113 \\
& 18130
\end{aligned}
$$ \& 0.15

0.18 \& $\begin{array}{ll}\text { ASYOS } & 0.20 \\ \text { ASY } 29 & 0.25\end{array}$ \& $\begin{array}{ll}\text { Cll } & 0.65 \\ \text { CRAI／05 } & 0.85\end{array}$ \& $0 \mathrm{Cl16}$ \& 0.50
0.38 \& \％TX500 \& 0.16 <br>
\hline 18131 \& 0.13 \& A8Y86 0.88 \& CRS1／40 0.45 \& OC19 \& 0.37 \& \％TX503 \& 0.17 <br>
\hline 18202 \& 0.23 \& Asz： 10.48 \& csin e－50 \& OC20 \& 0.85 \& ZTX531 \& 0.25 <br>
\hline 20371 \& 0.22 \& $\begin{array}{ll}\text { A } 2823 & 0.75\end{array}$ \& Csi013 3.18 \& 0 C 22 \& 0.50 \& \& <br>
\hline 2 G 381 \& 0.25 \& A1FY10 0．88 \& DD000 0.15 \& OC23 \& 0.80 \& INTEG \& TED <br>
\hline $2 \mathrm{G414}$ \& 0.92 \& AL109 1.80 \& DD003 0．15 \& OC24 \& 0.60 \& CIRCU \& <br>
\hline 2G417
2 N 404 \& 0.20 \& $\begin{array}{ll}\text { BC10\％} \\ \text { BC108 } & 0.10 \\ 0.10\end{array}$ \& DD00ti 0.18 \& $0 \mathrm{OC}_{2} 5$ \& 0.97 \& T401 \& 0.20 <br>
\hline $2 \mathrm{N69} \mathrm{\%}$ \& 0.15 \& $\begin{array}{ll}\text { 13C109 } & 0.10\end{array}$ \& $\begin{array}{ll}\text { DD0 } \\ \text { DD008 } & 0.48 \\ \end{array}$ \& OC26 \& 0.25 \& T401 \& 0.20 <br>
\hline 2N698 \& 0.40 \& $\mathrm{BCL13} 0.15$ \& （AD3 0.33 \& OC28 \& 0.60 \& T402 \& 0.20 <br>
\hline 2N706 \& 0.10 \& BC115 0.20 \& $\mathrm{aD}_{4} \quad 0.05$ \& 0－629 \& 0.60 \& 7403 \& 0.20 <br>
\hline $2 N 708$. \& 0.12 \& HC116 0.26 \& GD5 0.83 \& 0 CaO \& 0.60 \& T 10.4 \& $0 \cdot 20$ <br>
\hline 2N708 \& 0.15 \& BC118A 0.80 \& GD8 0.25 \& －0c35 \& 0.40 \& 740\％ \& 0.20
0.30 <br>
\hline $2 \mathrm{~N}^{2} 09$ \& 0.68 \& $\begin{array}{ll}\mathrm{BC1} 18 & 0.25\end{array}$ \& AD1 \& $\mathrm{OCP}^{0}$ \& 0.50 \& T40］ \& 0.30
0.30 <br>
\hline 2 N 1091 \& 0.88 \& BC121 0．20 \& GETIO： 0.80 \& OC36 \& 0.60 \& i 40 \& 0.30 <br>
\hline 2N1131 \& 0.25 \& 13C12： 0.20 \& GET103 0.22 \& OC 41 \& 0.25 \& 740\％ \& 0．20 <br>

\hline 2N1132 \& 0.25 \& ${ }^{3 \mathrm{C} 125} 00.88$ \& GET113 0.20 \& OC4： \& 0.80 \& | $1+49$ |
| :--- |
| $7+10$ | \& 0.45

0.20 <br>
\hline 2 N 1302 \& 0.18 \& $\mathrm{BCl}^{\text {B }} 60.85$ \& GET114 0.18 \& 0 OC 4 \& 0.40 \& i＋16 \& 0.20
0.23 <br>
\hline ${ }^{2} \mathrm{~N} 1303$ \& 0.18 \& $\mathrm{BC140} 0.55$ \& GETII5 0.45 \& $\mathrm{O}_{0} \mathrm{C} 4$ \& 0.17 \& 7411． \& 0．23 <br>
\hline ${ }^{2} \mathrm{~N} 1304$ \& 0.22 \& 112147

13148
0.15 \& GET116 0.50 \& $0 \mathrm{CH4M}$ \& 0.17 \& 48 \& 0．30 <br>
\hline 2N1305
2N1306 \& 0.22
0.25 \& $\begin{array}{ll}\text { BC148 } \\ 13 \mathrm{Cl} 49 & 0.13 \\ 0.16\end{array}$ \& $\begin{array}{ll}\text { GET120 } & 0.25 \\ \text { GET87\％} & 0.30\end{array}$ \& OCt5 \& 0.12 \& － \& 0.30 <br>
\hline 2 N 1307 \& 0.25 \& ${ }^{\text {HCLE }}$ \& $\begin{array}{ll}\text { GET875 } & 0.25\end{array}$ \& OCH i \& ${ }_{0}^{0.18}$ \& 7417 \& 0.30 <br>
\hline 2N1308 \& 0.25 \& ${ }^{13 \mathrm{Cl}}$ 58 0.12 \& $\begin{array}{lll}\text { GET } & 080 & 0.97\end{array}$ \& 0 O57 \& 0.60 \& －+20 \& 0.20 <br>
\hline 2 N 2147 \& 0.75 \& BCltio 0．63 \& （EET881 0.25 \& Oc5s \& 0.80 \& 1＋2．2 \& 0.48 <br>
\hline 2N2148 \& 0.80 \& $\begin{array}{ll}\text { BC169 } & 0.13\end{array}$ \& GET884 0.25 \& OC59 \& 0.65 \& 742：3 \& 0.48
0.48 <br>
\hline ${ }^{2} \mathrm{~N} 2160$ \& 0.80 \& $\begin{array}{ll}13 \mathrm{CY} 31 & 0.85\end{array}$ \& CET885 0.25 \& ${ }^{\text {OC6 }}$＋ \& 0.60 \& 7429 \& 0.48 <br>

\hline $$
\begin{aligned}
& \text { 2N2218 } \\
& \text { oNoong }
\end{aligned}
$$ \& 0.80

0.20 \& BCY3： 0.58 \& $\begin{array}{ll}\text { GEX44 } & 0.08 \\ \text {（EEX } 5 / 1 & 0.10\end{array}$ \& OC70 \& 0.12 \& － \& 0.50 <br>
\hline 2 N 2369.4 \& 0.15 \& $\begin{array}{ll}\text { BCY33 } & 0.25\end{array}$ \& （1EX9＋1 0．15 \& OC7： \& 0.20 \& 7436 \& 0.20 <br>
\hline 2N2444 \& 1.89 \& BCY34 0.30 \& （iJ3M 0.25 \& OC73 \& 0.30 \& $743:$ \& 0.48 <br>
\hline 2 N 2613 \& 0.88 \& HCY38 0.40 \& $(\mathrm{T} 44 \mathrm{M} 0.88$ \& 0 C 7 \& 0.30 \& 7．133 \& 0.70 <br>
\hline 2 N 2646 \& 0.45 \& BCY39 1．00 \& （iJ5．1 0．25 \& 0 CJs \& 0.25 \& 74.37 \& 0.65 <br>
\hline 2 N 2904 \& 0.20 \& BCY40 0.50 \&  \& Oc： 6 \& 0.25 \& 743\％ \& 0.65 <br>
\hline 2 N 2904 A \& 0.25 \& BCY42 0．25 \& HGi005 0.50 \& $0 \mathrm{C}^{7}$ \& 0.40 \& 7441 \& 0.20 <br>
\hline 2N2906 \& 0.20 \& BCY70 0．15 \& $188100.4 \quad 0.20$ \& 0 C 78 \& 0.20 \& 7411．1 \& 0.76 <br>
\hline 2 N 2907 \& 0.23 \& BCY71 0．20 \& Matiou 0.25 \& 0С79 \& 0.22 \& 744． \& 0.75 <br>
\hline 2N2924 \& 0.23 \& BCZ10 0．85 \& MAT］01 0.30 \& 0 C 81 \& 0.80 \& \％ \& 0.20 <br>
\hline 2 N 2925 \& 0.15 \& 15CZ1］0－50 \& Matien 0.25 \& 0c911） \& 0.20 \& 74.1 \& 0.20 <br>
\hline 2 N 2925 \& 0.10 \& BD121 0．65 \& Matiel 0.30 \& ос9im \& 0.20 \& 243 \& 0.20 <br>
\hline 2 N 3054 \& 0.50 \& $\begin{array}{ll}\text {［DD } 123 & 0.80\end{array}$ \& MJE520 0．87 \& Oc¢ ${ }^{\text {did }}$ \& 0.18 \& 2454 \& 0.20 <br>
\hline 2 N 3055 \& 0.75 \& BD124 0.75 \& MJE29j5 1.87 \& OC817 \& 0.40 \& T 4 \％ 10 \& $0 \cdot 20$ <br>
\hline 2N3702 \& 0.10 \& BDY11 1．62 \& MJE3055 0 －87 \& OC82 \& 0.25 \& 7470 \& 0.30 <br>
\hline 2 N 3705 \& 0.10 \& BFILS 0.25 \& NKT128 0.35 \& 008211 \& 0.20 \& 74 2 \& －0．80 <br>
\hline 2N3700 \& 0.28 \& BF117 0.50 \& NKT129 0.30 \& 0c83 \& 0.85 \& 3473 \& 0.40 <br>
\hline 2N3707 \& 0.12 \& BF167 0.26 \& NKT211 0.25 \& OC84 \& 0.25 \& 3474 \& 0.40 <br>
\hline 2N3709 \& 0.10 \& 4 Fl 1730.25 \& NKT213 0.26 \& OCl14 \& 0.88 \& 7475 \& 0.55 <br>
\hline 2N9710 \& 0.10 \& HF181 0．85 \& NKT214 0.15 \& OC12： \& 0.60 \& 74引 \& 0.45 <br>
\hline 2N3711 \& 0.10 \& BF184 0.20 \& $\begin{array}{lll}\text { NKT216 } & 0.37\end{array}$ \& $0 \mathrm{Cl23}$ \& ${ }_{0} .85$ \& 7480 \& 0.80 <br>
\hline 2N3819 \& 0.35 \& BF185 0.20 \& NKT 217
0.35 \& OC139 \& 0.25 \& $348: 3$ \& 0.87 <br>
\hline 2 N 5027 \& 0.68 \& BF194 0.17 \& NKT218 1.13 \& $\mathrm{OCl}_{40}$ \& 0.95 \& 7483 \& 1.00 <br>
\hline 2N5088 \& 0.83 \& BF190 0．15 \& NKT210 0．33 \& OClil \& 0.60 \& i484 \& 0.80 <br>
\hline 28301 \& 0.50 \& $1 \mathrm{FF196} 0.15$ \& $\begin{array}{ll}\text { NKT } 222 & 0.20\end{array}$ \& OC． 169 \& 0.20 \& 7486 \& 0.45 <br>
\hline 28304 \& 0.75 \& BF197 0.15 \& $\begin{array}{lll}\text { NKT224 } & 0.22\end{array}$ \& OC170 \& 0.25 \& 7490 \& 0.75 <br>
\hline 28801 \& 0.37 \& $1 \mathrm{FFB61} 0.28$ \& NKT251 0.24 \& OC171 \& 0.30 \& 7491．AN \& ${ }^{1} .00$ <br>
\hline 28702 \& 0.62 \& BFS94 0．28 \& NKT271 \& OC200 \& 0.40 \& 7492 \& 0.75 <br>
\hline AA129 \& 0.20 \& HFX12 0.20 \& $\begin{array}{ll}\text { NKT272 } & \text { 0．25 }\end{array}$ \& 0 C 201 \& 0.70 \& 7493 \& 0.75 <br>
\hline AAZ12 \& 0.80 \& BFX13 0．25 \& $\begin{array}{lll}\text { NKT273 } & 0.15\end{array}$ \& OC202 \& 0.80 \& 7494 \& 0.80 <br>
\hline AAZ13 \& 0.12 \& BFX29 0．26 \& NKT274
0 \& OC203 \& 0.40 \& 7493 \& 0.80 <br>
\hline AC107 \& 0.37 \& BFX30 0．25 \& NKT270

0 \& OC－204 \& 0.40 \& T +96 \& 1.00 <br>
\hline AC126 \& 0.20 \& BFX35 0．98 \& NKT277 0.20 \& OC205 \& 0.75 \& 7497 \& 6.25 <br>
\hline AC127 \& 0.25 \& BFX63 0.50 \& NKT278
0.26 \& OC206 \& 0.90 \& $i+100$ \& 2.50 <br>
\hline AC128 \& 0.20 \& ILFX84 0．25 \& NKT301 0.40 \& OC20－ \& 0.90 \& $7410{ }^{7}$ \& 0.50 <br>
\hline A0187 \& 0.25 \& BFX85 $\quad 0.30$ \& NKT304 0.76 \& OC460 \& 0.20 \& 74110 \& 0.80 <br>
\hline AC188 \& 0.25 \& BFX $86 \quad 0.25$ \& NKT403 0．75 \& OC470 \& 0.80 \& 74111 \& 1.45 <br>
\hline ACY17 \& 0.30 \& BFX87 0.28 \& NKT404 0.55 \& OCP7 \& 0.97 \& －4118 \& 1.00 <br>
\hline ACY18 \& 0.28 \& BFX88 0.20 \& NKT678 0.30 \& ORP1： \& 0.50 \& ¢ 4119 \& 1.80 <br>
\hline ACY18 \& 0.25 \& BFY10 1．00 \& $\begin{array}{llll}\text { NKTI3 } & 0.25\end{array}$ \& ORP60 \& 0.40 \& $7+121$ \& 0.80 <br>
\hline ACY20 \& 0.20 \& BFY11 1.25 \& $\begin{array}{cc}\text { NKTiJ3 } & 0.25\end{array}$ \& ORP61 \& 0.42 \& 74122 \& $1 \cdot 35$ <br>
\hline ACY21 \& 0.20 \& BFY17 0.25 \& $\begin{array}{lll}\text { NKT777 } & 0.38\end{array}$ \& ${ }_{819 \mathrm{~T}}$ \& 0.80 \& $7+123$ \& 2.70 <br>
\hline ACY2： \& 0.10 \& $\begin{array}{ll}\text { BYF18 } & 0.25\end{array}$ \& 07830 \& S． $\mathrm{Cl} 40^{\text {d }}$ \& 0.26 \& 74141 \& 1.00 <br>
\hline ACY27 \& 0.25 \& $\begin{array}{ll}\text { BFY19 } & 0.26\end{array}$ \& OA5 0.20 \& SFT308 \& 0.38 \& 74140 \& 1． 50 <br>
\hline ACY28 \& 0.17 \& BFY24 0.45 \& $\begin{array}{ll}0.6 & 0.12\end{array}$ \& \＄Ti22 \& 0.38 \& 74150 \& $3 \cdot 35$ <br>
\hline ACY 39 \& 0.50 \& BFY44 1.00 \& OA47 0.10 \& 8T7231 \& 0.63 \& 74151 \& $1 \cdot 10$ <br>
\hline ACY 40 \& 0.15 \& BFY50 0．22 \& OA70 0.10 \& ${ }^{9} \times 68$ \& 0.20 \& 74104 \& 2.00 <br>
\hline ${ }^{\text {ACY }} 41$ \& 0.15 \& BFY51 0．20 \& 0.4710 .10 \& 8×631 \& 0.30 \& －4150 \& 1.55 <br>
\hline ACY44 \& 0.25 \& 13FY59 0．22 \& 0.4730 .10 \& $8 \times 635$ \& 0.40 \& 74106 \& 1.56 <br>
\hline AD140 \& 0.50 \& ${ }^{\text {BFY53 }}$ \& 0474 0.10 \& SX 640 \& 0.50 \& 74157 \& 1.80 <br>
\hline AD149 \& 0.50 \& ${ }_{\text {BFY } 64} \quad 0.42$ \& 0.4790 .10 \& $8 \times 641$ \& 0.55 \& 74180 \& 4.10 <br>
\hline AD161 \& 0.87 \& BFY90 0．65 \& $0481 \quad 0.08$ \& $8 \times 6+2$ \& 0.80 \& 74174 \& $2 \cdot 00$ <br>
\hline AD162 \& 0.87 \& B8x27 0.50 \& $0 \mathrm{~A} 85 \quad 0.12$ \& SX644 \& 0.75 \& 7417\％ \& 1.35 <br>
\hline AF106 \& 0.80 \& BSX60 0.98 \& OA8B 0.15 \& 8X645 \& 0.75 \& 7416 \& 1.80 <br>
\hline AF114 \& 0.25 \& B8X 760.15 \& 0．490 0．08 \& V15／30P \& 0.50 \& － 4140 \& 1.95 <br>
\hline AF115 \& 0.25 \& BsY20 0．18 \& 0.49100 .07 \& V30／2011 \& 0.75 \& $\stackrel{+101}{ }$ \& 1.95 <br>
\hline AF116 \& 0.25
0.25 \& $\begin{array}{ll}\text { B8Y } 27 & 0.17 \\ \text { B8Y61 } & 0.60\end{array}$ \& $\begin{array}{ll}0.495 & 0.07 \\ 0.2000 & 0.07\end{array}$ \& $160 / 201$ \& 0.50 \& 74102 \& 2.00 <br>
\hline AF117 \& 0.28
0.62 \& $\begin{array}{ll}\text { B8Y51 } & 0.50 \\ \text { BSY95．A } & 0.12\end{array}$ \& $\begin{array}{ll}0.2200 & 0.07 \\ 0.202 \% & 0.10\end{array}$ \& ¢60／201P
$\mathbf{X}+101$ \& 0.75 \& －14193 \& 2.00 <br>
\hline AF119 \& 0.20 \& BSY90 0.12 \& OA2111 0－25 \& Xalol \& 0.10
0.18 \& 24194 \& 2.50 <br>
\hline AF124 \& 0.26 \& HT 102／500R \& 0 － $211 \quad 0.30$ \& － 4151 \& 0.15 \& T419． \& 1.85 <br>
\hline AF125 \& 0.20 \& 0.78 \& OAZ200 0．65 \& XA152 \& 0.16 \& T4195 \& 1.50 <br>
\hline AF126 \& 0.17 \& BTY4： 0.82 \& OAZ201 0－50 \& X
$\times 162$
$\times 161$ \& 0.16 \& \& <br>
\hline AF127 \& 0.17 \& BT $\times 79 / 100 \mathrm{H}$ \& 0 AZ 02 L 0．42 \& X 4161 \& 0.25 \& － \& 4.80
4.80 <br>
\hline AF39 \& 0.80 \& 0.75 \& 0 az203 0.42 \& X CBliz \& 0.25 \& －1199 \& 4.60 <br>
\hline AF178 \& 0.55 \& 131859／400 \& 0az204 0．30 \& XB101 \& 0.43 \& \multicolumn{2}{|l|}{\multirow[t]{6}{*}{Plies in sucketx luw profile 14 fin DlL 0.15 Ji pin Dll 0.17}} <br>
\hline AF179
AF180 \& 0.65
0.82 \& Byiou $\begin{aligned} & \text { a } \\ & 0.15\end{aligned}$ \& $\begin{array}{ll}\text { OAZ205 } & 0.42 \\ 0 \text { az206 } & 0.42\end{array}$ \& X B102 \& 0.10 \& \& <br>
\hline AF181 \& 0.42 \& $33 \mathrm{Y} 126 \quad 0.16$ \& ${ }^{0 .} \begin{array}{lll}0.20 \% & 0.47\end{array}$ \& X B 103 \& 0.25 \& \& <br>
\hline AF186 \& 0.40 \& $\mathrm{BY}^{\text {B }} 1270.17$ \& ${ }^{\text {OAZ } 208} \quad 0.32$ \& X X 113 \& 0.12 \& \& <br>
\hline AFY19 \& 1.18 \& BY18： 0.88 \& 0．AZ209 0.82 \& X $\mathrm{B}_{121}$ \& 0.48 \& \& <br>
\hline AFZII \& 0.60 \& $\begin{array}{ll}\text { BY213 } & 0.85\end{array}$ \& OAZ210 0．32 \& 21624 \& $0 \cdot 68$ \& \& <br>
\hline
\end{tabular}

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## TRANSFORMERS

[^6]

Recently I set about designing a system to enable me to remotely switch on and off two or more table lamps. I wished to be able to switch the lights on and off at any of three points regardless of the state of the switches at the other two.

This I decided could best be achieved by placing an electrically alterable bistable switch near the power point that fed the two lamps and running a twin flex lead from the bistable, connecting the three switches in parallel. Pushing on and releasing any switch would then change the bistable from on to off, or off to on depending on its history.

Since I had several suitable relays, and triacs are a little costly, I decided to use a relay to do the actual work of switching the lamps. The circuit I originally tried was the conventional type of bistable.


Fig. 1. Circuit diagram of the light control

Unfortunately, it had the disadvantage of requiring a fast edge to switch reliably, a faster edge than one could hope to regularly achieve with a mechanical switch. This meant, therefore, placing a monostable between the mechanical switch and the bistable. A rethink at this point produced the circuit shown in Fig. 1.

If TR1 is off and S1 is open, contacts RLA1 and RLA2 will be closed, point $y$ will be at zero potential and capacitor Cl will charge to -20 V through R1. If S 1 is then closed, point $y$ is reduced to the potential of point $x$, i.e. -20 V . The Zener diode D1 then conducts, resistor R3 limiting the current, and TR1 switches on. The relay is energised and the relay switches over.
The circuit will now take the form shown in Fig. 2, points $x$ and $y$ at $\frac{12 \times 20}{(3 \cdot 3+12)}$ negative volts (approximately -16 V ), which is just sufficient to maintain TR1 conducting. When S 1 is opened again, point $y$ goes to -20 V potential maintaining TR1 in conduction; capacitor C1 discharges through R1 to zero.
When S1 is pressed close, point $y$ is reduced instantaneously to zero, TR1 switches off and the relay contacts revert back to their original state.

Points $x$ and $y$ should now be at $\frac{3 \cdot 3 \times 20}{3 \cdot 3+12}$ negative volts (roughly -4 V ), insufficient to cause the Zener diode to conduct and TRI remains off. When SI is released (open) $y$ goes to zero and $x$ charges to -20 V through resistor R 1 . The cycle may now be repeated.

Because of the delay in charging capacitor C 1 the circuit will only change state when there is a minimum of two seconds delay between switching pulses, i.e. between S1 going off and coming on again. In certain circumstances this could be useful.

The delay could be reduced by reducing the values of R1 and R2 or C1. This is limited of course since C1 must not be allowed to charge or discharge through R2 in that instant when S1 is closed and the relay has not switched; in other words the minimum vales of $\mathrm{Cl}, \mathrm{R} 1$ and R 2 are primarily limited


Fig. 2. Effective circuit when S1 is closed
by the inertia of the relay. It should also be remembered that lower values of R1 and R2 will require greater transient power pulses from the supply.

It was found that the bistable operated satisfactorily with up to about 50 ohms between SI and point $x$. Line resistance to S 1 is therefore not important.
The switch pulse duration may be any length from a few hundred milliseconds (sufficient to switch the relay) to infinity.

The power supply is not critical and a bell transformer driving an unregulated voltage doubler was found satisfactory. Using a G.P.O. 1,000 ohm relay the circuit uses approximately 20 mA when switched on, and negligible current in the off state.

Any relay with double changeover contacts, other than very low coil resistance types, should work. The line voltage should be approximately the relay working voltage. The Zener diode should be $0.6 \times$ line voltage. Additional heavy duty contacts would be required for switching the lamps.
An $n p n$ transistor may be used in the circuit if preferred. In this case the Zener must be reversed, and the supply and electrolytic polarities changed.
P. King,

Withington,
Manchester.


THe circuit depicted in Fig. 1. wals designed to automatically control the running of two trains through a single-line section, both trains always travelling in the same direction as each other.

Referring to Fig. 2, points A and B are set for the inner loop. Whilst train 2 is travelling round the single-line loop train. 1 is held in isolating section $X$. Isolation of sections $X$ and $Y$ is controlled by the points blades at $B$. Track current is fed to either $X$ or $Y$ not both.

As train 2 passes position L , it interrupts a light beam focused on a light dependent resistor PCC1 (Fig. 1.) thereby triggering the desired sequence of switching operations. First point A changes to outer and then point B also changes to the outer rack.

When point B is set for the outer section, current is fed to section $X$, train $I$ starts up and proceeds around the single line section; the sense of points $B$ will cause train 2 to halt in section $Y$.
from the collector of TR1 switches on TR2 and TR3, which in turn energises RLA.

Relay RLA, however, remains closed for a very short time until capacitor Cl has charged up and consequently switched off TR2 and TR3. Relays RLA and RLB are wired in a master-slave relationship, for while RLA is closed a secondary set of its contacts is being used to charge up capacitor C2, via R6 and the RLAI contacts P and R. When RLA drops out contact $P$ makes contact with contact $Q$ (its normal state) and capacitor C2 discharges via R7 and TR4 base. This then switches on TR4 and TR5, but as before the pulse is momentary, until C2 has discharged sufficiently to reverse bias TR4.

After one cycle of operations has been completed, further triggering is not possible until after a delay (approximately 10 sec ) during which Cl discharges.

If desired, a variable resistor may be substituted for R2 to provide sensitivity control of the PCC1. The relays used were 12 V types. The contacts on RLA2 need to be of the break-before-make type, but the contacts RLA2 and RLB1 need only be simple on-off types; normally off.

The light source used was an ordinary flashlight bulb masked to shine through a $\frac{1}{4}$ in aperture into a 3in long cardboard tube housing PCCI at the far end.
J. Duffill,

Southam.


Fig. 1. Circuit diagram for the light operated model train controller. Note that C1 and C2 are electrolytics

As train I now passes position L it triggers off the same switching sequence as before, except this time the moving train is now on the outer loop and the points are therefore changed from outer to inner. By the time train 1 is halted in section $X$ train 2 is on the move again, we are back at the beginning of the sequence of operations and the whole cycle repeats.

The points are actuated by point ${ }^{\text {f }}$ motors which consist of a 2 -coil solenoid and an independently wired 2 -way switch that changes synchronously with the points motor, see Fig. 3. Since for reliable operation these motors need a relatively high current, it was arranged to operate them in sequence. They also require a one-shot pulse to avoid overheating the coils.

The pulses are supplied by sets of contacts on the relays RLA2 and RLBI, which close in sequence. The cross-wiring of the 2 -way switches on the points motors (Fig. 3) provides a flip-flop arrangement by which the closing of RLA2 causes point A to change to the sense determined by the setting of the switch S2 on the points motor M2. Similarly, when RLB1 closes, immediately on RLA2 dropping out, point B changes under control of the switch S1 on, motor MI.

The circuit Fig. 1 comprises two relay buffer stages and a light-operated circuit for triggering. Interruption of the light beam at L causes TRI to switch on because of the increased resistance of PCCI. Current


Fig. 2. Typical track layout


Fig. 3. Switch wiring arrangement

## FOR DEPTFORD RROADWAY, LOWDON, SE8 GGN

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MM12 $12 \mathrm{~V}, 250 \mathrm{~mA}+12 \mathrm{~V}, 250 \mathrm{~mA}$ MM $2020 \mathrm{~V}, 150 \mathrm{~mA}+20 \mathrm{~V}$. 150 mA L.T. $\quad 1,29$ plus 13p p . \& P .
L.T. ${ }^{\text {TT }} 6 \mathrm{VV}, 1.5 \mathrm{~A}-75 \mathrm{p}$ plus $18 p p$. \& $p$
 LT4 $12 \mathrm{~V}, 5 \mathrm{BA}, 61.32$ plus 30 p LT5 9-0-9V, 0.5A-75p plus $21_{p}^{\text {p. }}$ LT6 12-0-12V, 1A-95p plus. 26p $^{\text {p }}$ LT7 30-0-30V, IA-61.87 plus 30 p . Multi-tapped
MT30/2 0-12-15-20-24-30V, 2AMT60/1 0-5-20-30-40-60V. IA. MT60/2 $0-5-20-30-40-60 \mathrm{~V}, \mathrm{p}$. \& PA . $\quad \quad 2.95$ plus 34p p. \& p .
CTIO1 IA- 11.05 plus 26p p. \& p.
CTI02 2A- 11.30 plus $30 \mathrm{p} \mathrm{p} .\mathrm{\&} \mathrm{p}$.
CT/03 4A- 11.60 plus 30 p p. \& \&
Secondaries- $0-5-11-17 V$.

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50p rectifier-200 P.I.V., 2.0A , Transistor sockets-7p
Transistor sockets-7p
O.I. $1 . C$. sockers-14 pin, 20p $16 \mathrm{Pin}, 20 \mathrm{p}$
$\mathrm{N} 4001-50 \mathrm{p}$
N4001-50 P.I.V. $1 \cdot 0 \mathrm{~A} .6 \mathrm{p}$
$\mathrm{N} 4002-100$ P.IV. $1.0 A$
IN 4003 - 200 P.I.V., I.OA, 8 PD iN4004-400 P.IV., $1.0 A, 9_{p}$
IN $4005-600$ P.I.V., $1.0 A, 12 p$

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|  | R |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $1 \mu$ | 450 V | 19p | 1,000 | 25 V | 27p |
| $2 \mu \mathrm{~F}$ | 450 V | 20p |  | Sov | P |
| $4 \mu \mathrm{~F}$ | 350 V | 14p | 2.000 | 25 V | 39p |
| $8 \mu \mathrm{~F}$ | 450 V | 17p | 2.000 | sov | 53p |
| $16 \mu \mathrm{~F}$ | 450 V | 18p | 2,500 | 25 V | 45p |
| $25 \mu \mathrm{~F}$ | 25 V | 7 p | 2.500 | 50 V | 60 p |
| $25 \mu \mathrm{~F}$ | 50 V | 10p | 3,000 | 25 V | 48 |
| $32 \mu \mathrm{~F}$ | 450 V | 27p | 5,000 | 25 V | 60 |
| $50 \mu \mathrm{~F}$ | 50 V | 10p | 5,000 | 50V | $1 \cdot 10$ |
| $100 \mu \mathrm{~F}$ | $25 V$ | 10p | $8-8 \mu$ | 450 V | 18p |
| $100 \mu \mathrm{~F}$ | sov | $11 p$ | $8-16 \mu \mathrm{~F}$ | 450 V | 20p |
| $250 \mu \mathrm{~F}$ | 25 V | 14p | $16-16 \mu \mathrm{~F}$ | 450 V | 27 p |
| $250 \mu \mathrm{~F}$ | sov | 17p | $16-32 \mu \mathrm{~F}$ | 450 V | $63 p$ |
| $500 \mu \mathrm{~F}$ | 25 V | 18p | 32-32 $\mu \mathrm{F}$ | 450 V | 49 p |
| $500 \mu \mathrm{~F}$ | 50V | 25p | $50-50 \mu \mathrm{~F}$ | 350 V | 38p |

MINIATURE ELECTROLYTICS

| $1 \mu \mathrm{~F}$ | 63 V | 6 p | $47 \mu \mathrm{~F}$ | 16 V | 7 p |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $2 \cdot 2 \mu \mathrm{~F}$ | 63 V | 6p | $47 \mu \mathrm{~F}$ | 25 V | 6p |
| $3 \cdot 3 \mu \mathrm{~F}$ | 63 V | 6p | $68 \mu \mathrm{~F}$ | 16 V | 6 p |
| $4.7 \mu \mathrm{~F}$ | 63 V | 6p | $100 \mu \mathrm{~F}$ | IOV | 6p |
| $8 \mu \mathrm{~F}$ | 40 V | 7 p | $220 \mu \mathrm{~F}$ | 16 V | 7 p |
| $10 \mu \mathrm{~F}$ | 25 V | 6p | $330 \mu \mathrm{~F}$ | 16 V | $11 p$ |
| $10 \mu \mathrm{~F}$ | 64 V | $7 p$ | $470 \mu \mathrm{~F}$ | 10 V | $11 p$ |
| $16 \mu \mathrm{~F}$ | 40 V | $7 p$ | 1,000 ${ }^{\text {F }}$ | 16 V | 19p |
| $33 \mu \mathrm{~F}$ | 16 V | 6p | 1,500 F | 16 V | 23p |

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| :---: | :---: |
| Size Matrix | Matrix |
| 2tin $\times$ 3tin 22p | 16p |
| 2tin $\times$ 5in 24p | 25p |
| 3 in $\times 3$ 年in 24p | 25p |
| 3in $\times$ Sin 27p | 29p |
| 17in $\times 2$ inin ${ }^{\text {a }}$, 75p | 57p |
| 17in $\times$ 3sin 4 | 75p |
| Spot face cutter-39p <br> Pins-either size, packer of 3618 p |  |
| Edge connectors: |  |
| 24 way, 0.1 (2tin)-34p |  |
| 36 way, 0.1 (3xin)-44p |  |
| 16 way. 0.15 (21) in)-23p |  |
| 24 way, 0.15 (3i in)-34p |  |

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Car aerial |  |  | $\begin{array}{r} 14 p \\ 9 p \end{array}$ |  |  |  |
| Co-axialDil.N. 2 |  |  |  |  |  |  |
|  | in (speak | er) | 10p |  |  |  |
| D.I.N. 3 |  |  | ${ }^{13} \mathrm{p}$ |  |  |  |
| D.I.N. 4 |  |  | 14 p |  |  |  |
|  | in, $160^{\circ}$ |  | 13p |  |  |  |
| D.I.N. 5 | in, $240^{\circ}$ |  | 15p |  |  |  |
| D.i.N. 6 |  |  | 15 p |  |  |  |
|  | m unscre | ned | ${ }^{9} \mathrm{p}$ |  |  |  |
| Jack, ${ }^{\text {Jack, }} 3$ | $m$ screen |  | 10 p |  |  |  |
|  | $m$ unscre | ened | 8p |  |  |  |
| Jack, $3 \frac{1}{2}$ | $m$ screen |  | $12 p$ |  |  |  |
|  | nscreen |  | 12p | SO |  |  |
|  | Jack, tin screenedJack, stereo, uniscreened |  |  | 20p | Car |  |  |
|  |  |  |  | 20p | Caraerial Co-axial, | riace | 8p |
| Jack, stereo, screenedPhono, plastic top |  |  | 35p | $\begin{aligned} & \text { Co-axial, si } \\ & \text { Co-axial } \end{aligned}$ | riace ush | 8 p |
|  |  |  | ${ }_{12 p}$ | Co-axial, f $\text { D.I.N. } 2 \text { pi }$ | (speal | ${ }^{9 p}$ |
|  |  |  | 12 p | D.I.N. ${ }^{2} \mathrm{pi}$ | (speal | 9p |
| Wander, red or black <br> Banana 4 mm , red or black |  |  | 3 P | DIIN. ${ }^{\text {DIN }}$ | $180^{\circ}$ | p |
|  |  |  | 6 p | $\begin{aligned} & \text { D.I.N. } 5 \text { pi } \\ & \text { D.I.N. } 5 \text { pi } \end{aligned}$ | $\begin{array}{r} 180^{\circ} \\ 240^{\circ} \end{array}$ | 9p |
| LINE SOCKETS |  |  |  | $\text { Jack, } 2 \frac{1}{2} \mathrm{mn}$ |  | 10p |
| Caraerial |  |  | 14p | Jack, tin | nswitch | 5 p |
|  |  |  | $17 p$ | Jack, tin s | wirched | $17 p$ |
| Co-axial ${ }^{\text {D.I.N. } 2}$ pin (speaker) |  |  | $15 p$ | Jack, ster | O, switc | ed 24p |
|  |  |  | 16p | Phono, sin | le | $5 p$ |
| D.I.N. 5 pin, $180^{\circ}$ <br> DIN 5 pin $240^{\circ}$ |  |  | 16p | Phono, 2 on | n a strip | $7 p$ |
|  |  |  | 16p | Phono, 3 on | n a strip | 9p |
|  |  |  | 15 p | Phono. 40 | n a strip | 10p |
| Jack, tin screened |  |  | 49p | Wander, sin | ngle, red | or black 5p |
|  |  |  | 34p | Wander, | win stri | $7 p$ |
| Phono, placed metal |  |  | 14 p | Banana 4 m | m red, o | black 6p |
| CAPACITORS |  |  |  | $0.0027 \mu F$ | $500 \mathrm{~V}$ | S/M 15p |
| 2. 2 pF | 500 V | S/M | 719p | $0.0033 \mu \mathrm{~F}$ | 125 V | P.5. ${ }_{\text {cer }}$ |
| $3 \cdot 3 \mathrm{pF}$ | 500 V | S/M | 71 p | $0.0033 \mu \mathrm{~F}$ | 500 V | Poly. 6p |
| 5pF | 500 V | S/M | 71 p | $0.0033 \mu \mathrm{~F}$ | 1,000V | MDC 6p |
| 10 pF | $125 \vee$ | P.S. | 5p | $0.0036 \mu \mathrm{~F}$ | 500 V | S/M 15p |
| 10 pF | 500 V | S/M | $71 p$ | $0.0047 \mu \mathrm{~F}$ | 125 V | P.S. 9p |
| 15 pF | 125 V | P.S. | 5p | $0 \cdot 0047 \mu \mathrm{~F}$ | 500 V | Poly. 6p |
| 15 pF | 500 V | Cer. | 4p | $0.0047 \mu \mathrm{~F}$ | 500 V | \$/M 20p |
| 18pF | 500 V | S/M | 71p | $0.0047 \mu \mathrm{~F}$ | $1,000 \mathrm{~V}$ | MDC 6p |
| 22pF | 125 V | P.S. | 5p | $0.005 \mu \mathrm{~F}$ | 100 V | Mylar 3p |
| 22pF | 500 V | 5/M | 71 p | $0.005 \mu \mathrm{~F}$ | 500 V | Cer. 5p |
| 25 pF | 500 V | S/M | 7 p | $0.0068 \mu \mathrm{~F}$ | 125 V | P.S. $101 p$ |
| 27pF | 500 V . | Cer. | 4p | $0.0068 \mu \mathrm{~F}$ | 500 V | 5/M 30p |
| 33 pF | 125 V | P. 5 . | 5p | $0.0068 \mu \mathrm{~F}$ | 500 V | Poly. 6p |
| 33 pF | 500 V | S/M | 7 p | $0.0082 \mu \mathrm{~F}$ | 125 V | P.S. 10 fP |
| 39pF | 500 V | S/M | 7 f | $0.0082 \mu \mathrm{~F}$ | 500 V | S/M 30p |
| 47pF | 125 V | P.S. | ${ }^{5 p}$ | $0.01 \mu \mathrm{~F}$ | 18 V | Disc 4p |
| 47pF | 500 V | Cer | $4 p$ | $0.01 \mu \mathrm{~F}$ | 125 V | P.S. $101 p$ |
| SOpF | 500 V | 5/M | 710 | $0.01 \mu \mathrm{~F}$ | 160 V | Poly. 4p |
| 56 pF | 500 V | S/M | 7 p | $0.01 \mu \mathrm{~F}$ | 250 V | M.F. ${ }^{\text {Pp }}$ |
| 68 pF | 125 V | P.S. | 5p | $0.01 \mu \mathrm{~F}$ | 400 V | Poly. 3p |
| 68 pF | 500 V | S/M | $7 \mathrm{7p}$ | $0.01 \mu \mathrm{~F}$ | 500 V | Cer. 5p |
| 75 pF | 500 V | S/M | $71 p$ | $0.01 \mu \mathrm{~F}$ | 500 V | S/M 30p |
| 82pF | 500 V | S/M | 7 p | $0.01 \mu \mathrm{~F}$ | 600 V | MDC 7p |
| 100 pF | 125 V | P.S. | 5p | $0.01 \mu \mathrm{~F}$ | 1,000V | MDC 9p |
| 100 pF | 500 V | S/M | 7 p | $0.015 \mu \mathrm{~F}$ | 160 V | Poly. 3p |
| 100 pF | 500 V | Cer. | $S_{p}$ | $0.015 \mu \mathrm{~F}$ | 400 V | Poly. 3p |
| 120 pF | 500 V | 5/M | 7 p | $0.02 \mu \mathrm{~F}$ | 100 V | Myiar 3p |
| 150 pF | 125 V | P.S. | 5p | $0.022 \mu \mathrm{~F}$ | 18 V | Dise 5p |
| 150 pF | 500 V | S/M | 7 P P | $0.022 \mu \mathrm{~F}$ | 250 V | M.F. 3p |
| 150 pF | 500 V | Cer. | 5p | $0.022 \mu \mathrm{~F}$ | 400 V | Poly ${ }^{3 p}$ |
| 180 pF | 500 V | S/M | 7 7p | $0.022 \mu \mathrm{~F}$ | 600 V | MDC 7 1p |
| 200pF | 500 V | S/M | 7 p | $0.022 \mu \mathrm{~F}$ | 1.000 V | MOC 10p |
| 220 pF | 125 V | P.S. | 5p | 0.033 $\mu \mathrm{F}$ | 250 V | M.F. 4p |
| 220 pF | 500 V | Cer. | 5p | $0.033 \mu \mathrm{~F}$ | 400 V | Poly. 4p |
| 250 pF | 500 V | S/M | ${ }^{8 p}$ | $0.047 \mu \mathrm{~F}$ | 12 V | Disc 6p |
| 270 pF | 500 V | Cer. | 5 p | $0.047 \mu \mathrm{~F}$ | 160 V | Poly. 3p |
| 300 pF | 500 V | S/M | ${ }^{8 p}$ | $0.047 \mu \mathrm{~F}$ | 250 V | M.F. 3 P |
| 330 pF | 125 V | P.S. | 5p | $0.047 \mu \mathrm{~F}$ | 400 V | Poly. 4 p |
| 330 pF | 500 V | S/M | 8 p | $0.047 \mu \mathrm{~F}$ | 600 V | MDC 8p |
| 390 pF | 500 V | 5/M | ${ }^{8 p}$ | $0.047 \mu \mathrm{~F}$ | 1.000 V | MDC 10p |
| 470 pF | $125 V$ | P.S. | $5 p$ | $0.1 \mu \mathrm{~F}$ | 30 V | Disc 6p |
| 470 pF | 750 V | Diss | $5 p$ | $0.1 \mu \mathrm{~F}$ | 250 V | M.F. 4p |
| 500 pF | 500 V | S/M | 8 p | $0.1 \mu \mathrm{~F}$ | 400 V | Poly. 5p |
| 560 pF | 500 V | 5/M | ${ }^{8 p}$ | $0.1 \mu \mathrm{~F}$ | 600 V | MDC 10p |
| 680 pF | 125 V | P.S. | 6p | $0.1 \mu \mathrm{~F}$ | $1,000 \mathrm{~V}$ | MDC 14p |
| 680pF | 500 V | S/M | 8 p | $0.15 \mu \mathrm{~F}$ | 250 V | MF. 5p |
| 820pF | 500 V | S/M | 8 p | $0.22 \mu \mathrm{~F}$ | 160 V | Poly. 6p |
| $0.001 \mu \mathrm{~F}$ | 100 V | Mylar | 3p | $0.22 \mu \mathrm{~F}$ | 250 V | M.F. 5p |
| $0.001 \mu \mathrm{~F}$ | 125 V | P.s. | 6p | $0.22 \mu \mathrm{~F}$ | 400 V | Foil 10p |
| $0.001 \mu \mathrm{~F}$ | 400 V | Poly. | 3p | $0.22 \mu \mathrm{~F}$ | 1.000 V | MDC 15p |
| $0.001 \mu \mathrm{~F}$ | 500 V | S/M | 10 p | $0.33 \mu \mathrm{~F}$ | 250 V | M.F. ${ }^{\text {Pp }}$ |
| $0.001 \mu \mathrm{~F}$ | 500 V | Cer. | 5p | $0 \cdot 47 \mu \mathrm{~F}$ | 250 V | M.F. 8p |
| $0.001 \mu \mathrm{~F}$ | 1.000 V | MDC | 6p | $0.47 \mu \mathrm{~F}$ | 400 V | Foil 15p |
| $0.0015 \mu \mathrm{~F}$ | 400 V | Poly | ${ }^{3 p}$ | $0.47 \mu \mathrm{~F}$ | 1.000 V | MDC 25p |
| $0.0015 \mu \mathrm{~F}$ | 500 V | S/M | 10p | $1.0 \mu \mathrm{~F}$ | 250 V | M.F. 15p |
| 0.0015 $\mu \mathrm{F}$ | 500 V | Cer. | 5p |  |  |  |
| $0.0018 \mu \mathrm{~F}$ $0.002 \mu \mathrm{~F}$ | 500 V | S/M | 10p | Note: |  |  |
| $0.002 \mu \mathrm{~F}$ | 100V | Mylar | 3 p | S/M = sil | ver mica | $1 \%$ tol. |
| $0.002 \mu \mathrm{~F}$ | 500 V | Cer. | 5p | P.S. $=$ po | lystyrene | $2 \frac{1}{4} \mathrm{sol}$. |
| $0.0022 \mu \mathrm{~F}$ | 125 V | P.S. | 6p | MOC | c. ration | = 300V. |
| $0.0022 \mu \mathrm{~F}$ | 500 V | S/M | 10p | M.F. $=$ | ullard m | in. foit. |
| $0.0022 \mu \mathrm{~F}$ | $1,000 \mathrm{~V}$ | MDC | 6p | Cer. $=$ c | ramic. |  |

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$500 \mu \mathrm{f} 25$ volt
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$1,000 \mu \mathrm{f}$
25
$2,000 \mu \mathrm{val}$
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volt
p $2,500 \mu \mathrm{\mu} 50$ volt
$400 \mu f 40$ volt
$125 \mu f 4$ vole
$125 \mu f 4$ volt
$400 \mu f 6.4$ volt
$320 \mu \mathrm{f} 10$ volt

| $16 \mu \mathrm{f}$ |
| :--- |
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Two flash units from Honeywell

## SOLDERING PACK

Certainly an ideal kit to present as a gift to anyone, particularly at this time of the year, the new Adcola Products soldering kit contains all the essentials for simple soldering jobs. The kit contains an Invader soldering iron. a stand. solder wire. two spare soldering tips and an instruction card providing useful hints on successful soldering.

The soldering iron is thermally controlled to provide a constant heat level for ease of soldering. The standard soldering tip provided is ${ }_{3}^{3}$ in diameter, but two replacement tips of tin and tin diameter are included to provide complete versatility for a 25 W iron. The tips are mere!y inserted in the collar at the end of the iron to convert it from one task to another.

The stand provided with the kit features an integral sponge for cleaning excess solder from the tips and two holes to contain the spare tips. A spring holder is mounted on the base of the stand at an angle of about 45 degrees to take the iron when not in use.
The kit costs $£ 3.99$ and all the components are contained in a plastics tray featuring compartments for each component for safety in transit. The kit should be available from most d.i.y. shops, ironmongers or direct from Adcola Products Ltd.. Adcola House. Gauden Road, London. S.W. 4.

## EMERGENCY LANTERN

With the forecast that power cuts are again probable this winter. it is wise to plan ahead for the eventuality and to have some form of emergency lighting handy.


Items mentioned in this feature are usually available from electronic equipment and component retailers advercising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.


Adcola Products
Soldering pack
A useful lamp is the "Safety Lantern" available from Harris Marketing. It is claimed to be virtually unbrcakable, having a safe plastic dome which gives a 360 degree light and is powered by a 6 V 996 battery.

It is particularly suitable around the home where there may be children or elderly people, for whom candles and other naked lights would be a danger. but it also makes a useful car, caravan or camping light.

Available direct from Harris Marketing. 16 Hillcroome Road, Sutton, Surrey, the lamp costs $£ 1.85$ plus 15 p post and packing.

## PHOTOGRAPHIC

## EQUIPMENT

Famous for their work in the computer field Honeywell have announced their entry into the photographic world with a range of electronic flash equipment. 35 mm slide projectors and darkroom equipment.
Available in this country throug? Photo-Science Ltd. the range of equipment includes automatic flash units and slave flash units.

Also included in their range of products are four automatic selffocusing projectors with a special built-in preview screen. These models enable the projectionist to see each slide on a small screen before it is projected to the audience.
Two of the flash units represent Honeywell's latest technological advancements in this field, which they call "strobo-eye" remote sensing. The sensor device is completely separated from the flash so that the light can be controlled from the camera or wherever the photographer desires.


## Instant copying kit by Polaroid

These two flash units are described by Honeywell as the only system that will give accurate antomatic exposure control of oncamera. off-camera and wall- or ceiling-bounce flash lighting.

Further particulars of the complete range of Honcywell pholographic equipment can be obtained from Photo-Science Itd., Charfleet Road, Canvey Island. Essex

## INSTANT COPY KIT

An inexpensive instant copy kit and accessories. providing a complete photographic copying system. is being introduced by Polariod (UK) Lid. Both the copy kit and accessory unit are designed for use with any of three Polaroid instant picture cameras.
ldeal for making copies from a master design of printed circuit patterns to any required size. The Polaroid $105-1$ kit gives a $z^{2} 1$ reduction. The kit consists of a copy stand. close-up lens in mount. blue filter (for colour film used with tungsten light). polarising filter (for flash only). Colorpack so camera and a carrying case.

The complete kit costs $£ 38.95$ : without a camera the kit retails at $£ 24$

The 106 Instant Copy Kit accessory unit. comprising a copy stand and close-up lens for $\frac{1}{3}$ : I copies. will retail for $\ddagger 1500$

## EXTENSION LEADS

It has been brought to our attention that the price quoted for the Portapower cable reels mentioned in our September issue does not include the cost of cable.

We apologise for any inconvenience this may have caused.

# Repdone A SELECTON RHOM OUR POSTBAG 

Correspondents wishing to have a reply must enclose a stamped addressed envelope. We egret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

## Correct perspective

Sir-Mr Hickman (Readoul, November issue) has raised the question of component cost in Australia, a point certainly omitted from, the first Reporl from Austratia. Detailed costings of parts were not quoted. as indeed many other aspects of home construction were omitted, due mainly to space considerations; it was intended only as an introduction.

However. since the subject has been raised. the question of component cost must be placed in the correct perspective. In Sydney and Melbourne, the BC108, if we are to use this device as an example, can be bought at the rate of 6 for $\$ 2.00$ ( 16 p each) including sales (purchase) tax. The figures as quoted by Mr Hickman may be applicable to South Australia, but it is indeed a half-truth to suggest that his situation applies throughout Australia.

The cost of components generally must, in any case, be viewed against the background of higher salaries here. It has been stated that. for many professional occupations, salaries are very roughly 50 per cent higher than in the U.K

1 would agrer with your correspondent's comments regarding Australian conditions, particularly those
in respect of the somewhat overglamorised publications for which Australia House is now famous.

Finally. Mr Hickman is under the illusion that I am Australian, and tend to avoid comparisons. I myself migrated from England some five years ago.
J. Waldie.

Tamarama. N.S.W.,
Australia.

## The defender

Sir-I would like to come to the defence of your constructional features in general and R. W. Coles "P.E. Digi-Cal" in particular.
I doubt whether I will ever build the Digi-Cal. but the detailed way in which its operation is explained by the author has added measurably to my understanding of the logic functions involved. If such features are to be excluded because of a cost disadvantage compared with custom built devices. we amateurs will be deprived of much needed design know-how.

I have to confess that I jumped the gun by buying the Minitrons before the total cost of parts was revealed. Doubtless they will come
in handy one day-but it might be an idea if project-authors revealed the approximate cost of components at the outset and saved we impecunious constructors the embarrassment of having to abandon a project after buying a few of the items needed.

## J. A. Burness,

Herts.

## Valuable learning

Sir-I noticed a number of letters in the October issue of P.E. that objected to the Digi-Cal on monetary and aesthetic grounds. The question was raised, why should you publish an article outlining the construction of a calculator when the basic i.c. is now available.

I can certainly construct a very simple calculator around one of the many calculator integrated circuits that are now available. but this teaches me nothing new about digital electronics. Even though I may not want to build the Digi-Cal. I have certainly been able to follow along with the author and now 1 have a belter understanding of how these devices work. Maybe I have also picked up some new ideas on how to do things such as keyboard multiplexing and readout demultiplexing. That is as valuable as learning the basic operations of a calculator.

Many electronic hobbyists here in the USA are all too ready to put together a kit that teaches them nothing and tests only their ability to follow directions. Most of the electronic publications here cater to these individuals and that is why 1 subscribe to P.E. Keep up the good work.

Jonathan T. Titus.
President.
Titus Labs.
Blacksburg. USA.

## BACK NUMBERS WANTED

Anyone who can supply the undermentioned are asked to communicate directly with the reader.

```
March-April 1971
Mr R. Brown. P.O. Box 926, Chingola. Zambia,
July 1969
MrK. Farrance, Charity Coctage, High Rougham,
Bury St Edmunds, Suffolk.
Buryruary 197!
Mr D.J. Hitchcock, 38 Deakin Sereet. Hampton
3189. Melbourne, Vicroria, Australia.
December 1968
Mecember Mas8 Robson. 8 Withipoll Sereec, Ipswich.
Suffolk.
February, March. May 1971
February, March, May 197I
Sarawak.
January 1988
Mr G. Hart. 41 Waterlow Road, London N195NJ.
Mugust 1966
Ausust (986 Rume, 93 Sinclair Drive, Glasgow
G42 9PU.
December 1970 zo March 1971
December 1970 to March 197|, Erington Park,
New Marske, Redcar. Teesside.
New Marske, Redcar. Teess
May 1960 to March 1970
Barry,Glamorgan.
April }197
MrM. P., Baker. 14 Cabell Road, Park Barn Estate,
Guildford, Surrey.
March-April 1971
```

September 1966, January, June, July, Mrgust Stuck, 28 Chancelot Road. Abbey Wood, London SE2 ON
July 1972
MrP. J. Joyce. 21 Eggingron Courc. Loughborough University, Leicestershire.
December 1970 to March 1971
MrA. Jones. Tyddyn Slaters, Pontrus. Caernarvon

## North Wales

November 1971 to March 1972
Mr C. Wright, 98 Swallow Road. Ipswich. Suffolk

## July 1969

Mr M. Armitage, Oakendean, 12 Shanter Way. Alloway. Ayrshire
Mayto October, December 1969
Mr W. J. Miller, Dailard, 13 Adler Road. Glasgow 53.
November 1970 to March 1971
Mr J. Maplesden. 47 Oldford Crescent, Acklam, Middlesbrough. Teesside. TS 5 7EH.
January-February 1971
Mr C. E. R. Flynn, A.P.C. Ltd., Dungeness Power Station, Romney Marsh, Kent.
May 1972
May 1972 . Tawse, 25 Welland Road, Edith Weston
Oakham, Rutiand, LEIS BJE.

We regret that back numbers of Practical Electronics can no longer be supplied. We will try to publish announcements of readers' requirements (without a guaranteed date) free of charge.

May 1969 to February 1970
Mr S. P. Ganecki, 17 Belmont Close, Churchbridge, Cannock, Šaffordshire.
November-December 197!
Mr H. D. Westwood, 12 Cairn Walk. Erumlin Co. Antrim, N. Ireland

## June 1969

Mr J. E. Marshall, 4 Dellcott Close. Welwyn Garden City, Herts
Mpril-May 1971 Road, Kingston.
November 1971
Mr G. C. Street, 21 Linton Crescent. Hastings Sussex.
July 1970 to May 1971
MrR. Konarek, 573 Daws Heath Road. Hadleigh, Benfleet. Essex.
May, June, July, Sept., Oct., Nov., Dec, 1969. January to Aprif. August 1970
Mr J. Shapero, 64 Tixall Road. Birmingham, M2 28 ORS.
Apris-July 1971
Mr K, McKegney, 21 Queen Street, Chesterfield Derbyshire. 540 45F.

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## SPECIFICIATIONS

Number of iransistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz .
Sensitivity: $7 \mu \vee$ for lock-in ovel full deviation.
Squelch leval: typically $20 \mu \mathrm{~V}$.
Signal to noise ratio: $\pm 65 \mathrm{~dB}$.
Audlo frequancy response: $10 \mathrm{~Hz}-15 \mathrm{Khz}$ ( $\pm 1 \mathrm{~d}$ ).
Total harmonic distortion: $0.15 \%$ for 30\% modulation.
Stereo decoder operating level: $2 \mu \mathrm{~V}$.
Crosstalk: 40dB.
Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. max. (typically $2 \times 50 \mathrm{mV}$. stereo)
Operating voltage: $25-30 \mathrm{~V} D \mathrm{C}$ at 100 mA . Indicators: Stereo on: tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.

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Integrated circuit
high fidelity amplifier


Having introduced Integrated Circuits to hi-fi constructors with the IC.10. the first lime an IC had ever been made available for such purposes. we have followed it with an even more efficient version. the Super IC. 12. a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up. F.M. radio or small P.A. set up. etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible. It is the equivalent of a 22 tran-
sistor circuit contained within a 16 lead DIL package. and the finned heat sink is sufficient for all requirements. The Super IC. 12 is compatible with Project 60 modules which would be used with the $\mathbf{Z . 5 0}$ and $\mathbf{Z . 3 0}$ amplifiers. Complete with free manual and printed circuit board.

## SPECIFICATIONS

Output power: 6 watts RMS continuous ( 12 watts peak). 6-8 . Frequency Response: 5 Hz to $100 \mathrm{KHz} \pm 1 \mathrm{~dB}$. Total Harmonic Distortion: Less than $1 \%$. (Typical $0.1 \%$ ) at all output powers and frequencies in the audio band (28V). Load Impedance: 3 to 15 ohms. Input Impedance: 250 Kohms nominal. Power Gain: 90 dB ( 1.000 .000 .000 times) after feedback. $90 d 8$ ( 1.000 .000 .000 times) after feedback.
Supply Voltege: 6 to 28 V . Quiescent current: 8 mA at 28 V . Size: $22 \times 45 \times 28 \mathrm{~mm}$ including pins and heat sink.
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With FREE printed circuit board and 40 page manual.
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Project 605 is one pack containing: one P25 two 230 's. one Stereo 60 and one Masterlink. This new module contains all the input sockets and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting.
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passed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at $15 \mathrm{w}(8 \Omega)$ and all lower outputs. Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that Z .50 s and Z .30 may be used in a far wider range-of applications.
SPECIFICATIONS ( 2.50 units are interchangesb/e with $Z .30$ s in all applicetions).- Power Outputs: 2.3015 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts.
2.50 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.

Frequancy response: 30 to $300,000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. Distortion: $0.02 \%$ into 8 ohms. Signal to noise ratio; better than 70 dB unweighted. Input senaitivity: 250 mV into 100 Kohms (for 15 w into $8 \Omega$ ). For speakers from 3 to 15
ohms impedance. Size: $14 \times 80 \times 57 \mathrm{~mm}$.
 hms impedance. Size: $14 \times 80 \times 57 \mathrm{~mm}$


## Stereo 60 Pre-amp/control unit

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Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount
SPECIFICATIONS-Input sensitivities: Radio - up to 3 mV . Mag. p.u. 3 mV : correct to R.I.A.A. curve $\Psi^{11 \mathrm{~dB}}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u. - up to 3 mV : Aux - up to 3 mV . Output: 250 mV . Signal to noise ratio: better than 70dB. Channel matching: within 1 dB . Tone controls: TREBLE +12 to -12 dB at 10 KHz : BASS +12 to -12 dB at 100 Hz . Front panal: brushed aluminium with black knobs and controls. Size: $66 \times 40 \times 207 \mathrm{~mm}$.


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$£ 5.98$
For use between Stereo 60 unit and two $Z .30$ s or $Z .50$ s. The unit is very easify mounted and is unique in that the cut-off frequencies are continuously variable. As attenuation in the rejected band is rapid ( $12 \mathrm{~dB} /$ octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. There are two filter sections - rumble (high pass) and scratch (low pass). H.F. cut-off ( -3 dB ) variable from 28 KHz to 5 KHz . L.F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at 1 KHz ( 35 V . supply) $0.02 \%$ at rated output. Operating voltage from 15 to 35 V . Current 3 mA . Size: $66 \times 40 \times 90 \mathrm{~mm}$.

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Typical Project 60 applications

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| 25W. RMS continuous sine wave stereo amp. using lơw efficiency (high performance) speakers | $\begin{aligned} & 2 \times 2.30 \mathrm{~s}, \text { Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U.. F.M. Tuner, Tape Deck, etc. | £26.90 |
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| F.M. Stereo Tuner (£25) \& A.F.U. (£5.98) may be added as required. |  |  |  |
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| We can now offer a vast range of |  |  |  | SN7440 | 1.851 .70 | 8N74120 | 1.060 .98 |
| Motorola IC"s at industrial distributor |  |  |  | SN7446 | 2.00 1.60 | 8N74121 |  |
|  |  |  |  |  |  |  |  |
|  |  |  |  |  | 1.0 | SN7 |  |
| MC724P | $\begin{aligned} & 0.67 \\ & 1.24 \end{aligned}$ | 20.121 |  |  | Price ${ }^{\text {announ }}$ |  | 0.900 .80 |
| MC1303L |  |  |  |  |  | SN7 |  |
| FAIRCHILD (RTL) |  |  |  |  | 0.20 | SN741 | 1.191 .05 |
|  |  |  |  | SN | . | SN74153 | 1.1 |
|  |  |  |  | 8N7404 | 0.200 .18 | SN74154 1 | 1.89 1.75 |
|  |  | 37p |  | SN7460 | 0.200 .18 | SN74150 1 | 1.551 .85 |
| Data slieet 1210 |  |  |  | SN7470 | 0.300 .88 | SN74156 | 1.801 .15 |
|  |  |  |  | 8N74 | 0.800 .8 | SN74157 | 1.801 .15 |
| LINEAR |  |  |  |  |  |  |  |
|  |  |  |  | SN7475 | 0.450 .85 | SN74161 8 | 8.80 2.50 |
| Op. Ampr. |  |  |  | SN7476 | 0.440 .39 | SN74162 2 | 2.602 .50 |
| L702A T05 |  |  | $2 \cdot 60$ | 8N7480 | 0.750 .70 | SN74163 3 | 3-58 |
| $\begin{array}{ll}\text { L702C } \\ \text { L708C } & \text { T05 } \\ \text { T05 }\end{array}$ |  | 0.70 | 2 | SN7481 | 1.251 .07 | 8N74184 2 | 2.261 .64 |
|  |  | 0.67 | 0.52 | SN7483 | 0.870 .81 | SN74165 | 4.003 .85 |
| - | TO5 |  | 0.31 | SN74 |  | SN74106 |  |
|  | DIL | 81 | 0.27 | SN |  |  |  |
| L710CL 710 C | T05 | $77 \quad 0.48$ | 0.87 | SN7486 | 0.450 .40 | SN74174 2 | 2.001 .74 |
|  | DIL | 0.40 | 0.85 | SN7488 | 12.4011 .80 | SN74175 1 |  |
| L711C |  | 0.44 | 0.39 | SN7489 | $4 \cdot 97$ | SN7417 |  |
|  |  | 0.40 | 0.85 | SN7490 | 0.80 0.75 | SN741 | 10 |
| $\begin{aligned} & \mathrm{L} 716 \\ & \mathrm{~L} 723 \mathrm{C} \end{aligned}$ | T05 | ${ }^{1.75}$ | . 80 | 8N7491 | 1.201 .00 | ${ }_{\text {8N74178 }} 1$ | 1.801 .55 |
|  |  |  | -80 | SN7492 | 0.750 .70 | SN74179 1 | 1.801 .55 |
|  |  | 38 | 0.34 | SN7493 | $0.75 \quad 0.70$ | SN74180 | 1.371 .20 |
|  | DIL | 0.36 | 0.38 | SN7494 | 0.84 | SN74131 ${ }^{6}$ | 8.50 6.15 |
| LM74]CN DIL |  |  |  | SN7496 |  |  |  |
| 8N72700P DIL |  | $45 \quad 0.40$ | 0.85 | SN | ${ }_{6.25}$ | SN |  |
| GENERAL |  | TOSHIBA |  | 8N74100 |  | SN741 | 80 |
|  |  | SN74104 | $\frac{1}{1.59}$ | SN74193 | 1.90 1.80 |
| ELECTRIC |  |  |  | 20 WACE |  | ${ }^{8 N N 74107}$ | 0.480 .38 | ${ }_{\text {8N74193 }}$ |  |
|  |  | SN74109 | 1.080 .93 |  |  | 8N74196 | 1.801 .88 |
|  |  | TH9013P | 33.50 | 8N7410 | 0.800 .70 |  |  |
|  |  | PREAMP |  | SN74111 | 1.571 .38 | SN74197 1 | 1.80 1.88 |
| $\begin{array}{ll}\text { PA239 } & \text { 2.81 } \\ \text { PA246 }\end{array}$ |  | Limited nuıber |  | SN74118 | 1.00 0.78 | SN74198 | 4.00 8.75 |
|  |  | 8N74119 | 1.921 .68 | SN74199 4 | .00 3.75 |
| $\begin{aligned} & \text { PA264 } \\ & \text { PA2 } 265 \end{aligned}$ |  |  |  |  |  |  | 0 |  |  |
|  |  | SN74800N |  |  |  | SN74810N | $1 \cdot 10$ |
|  |  | MULLARDLINEAR |  | 8N7480 | 1.38 | SN74520N | $1 \cdot 10$ |
|  |  |  |  | SN74864 | N 1.10 | 8N74S65N | 1.10 |
| SIEMENS |  |  |  | 8N748401 | N 1.2 | 8N74S140N | 1.70 |
|  |  | A241 | 80 | 9N748112 | 2N 2.88 | 8N748114N | 2.86 0.80 |
|  |  | 242 | 4.00 | SN74L00 | ON $\quad \begin{aligned} & 0.54 \\ & 0.60\end{aligned}$ | gN74LO2N | 0.60 8.15 |
| 40 |  | 263 | 0.70 | SN74L |  | $\begin{aligned} & \text { 8N74L47N } \\ & \text { SN74L74N } \end{aligned}$ |  |
| TAAAS1TAA485 | 0.880.78 | 293 | 0.80 |  | 10 | SN74L93N | 08 |
|  |  | 300 | 2.00 |  |  |  |  |
|  |  | 310 | 1.80 |  |  |  |  |
|  |  | 320 | 0.65 | Type | $1-2425$ |  | -99 |
| PLESSEY |  | 350 | 8.10 |  |  |  |  |
|  |  |  | 435 | 0.78 | CA3000 | 2.28 | ${ }_{\text {CA }}$ | 2.08 |
|  |  | 521 | 0.57 | CA3 |  | CA3045 | 1.22 <br> 0.868 <br> 0.71 <br> 108 |
|  |  | 522 | 1.50 | CA3004 | 2.26 1-86 | CA3047 | 1-59 1.81 |
|  |  | 530 | . 95 | CA300 | 1.50 | CA3047A | 2.92 2.41 |
|  |  | 570 | 2.60 | CA3006 | 3.472.86 | CA3048 | 2.341 .98 |
| ${ }^{\text {8L6 }}$ 821C | 60 | 700 | 3.50 | CA3007 |  | CA3049 | 1.96 1.68 |
| ${ }^{\text {BL7 }}$ 21C |  | 811 | 4.45 | CA33 |  | $\mathrm{CA}^{\text {CA }}$ | $\begin{array}{ll}2.28 \\ 1.78 \\ 1 & 1.48\end{array}$ |
| 8L702C |  | AB101 | 0.971 | CA3010 | 0.56 0.47 | CA4052 | 1.801 .48 |
| DATA BELGETS$17 \ddagger p$ |  | Tadi00 | 1.97 | CA3010 | 0.690 .57 | СА 30 ̃3 | 0.620 .48 |
|  |  | Tadilo | 1.971 | CA3011 | 0.928 | CA3054 | 9.98 |
|  |  |  | CA3012 | 1.111 .00 | CA3058 | 8.88 8.74 |
| SGS |  |  |  | ${ }_{\text {CA3014 }}$ | ${ }_{1}^{1.52} 11.0$ | CA3060 ${ }^{\text {Ca }}$ |  |
| $\begin{array}{ll} \hline \text { TAA } 681 \mathrm{~B} & 81.88 \\ \text { TAA } 621 & 82.08 \\ \hline \end{array}$ |  |  | $\begin{aligned} & \text { TAA } 700 \text { e8.50 } \\ & \text { TBA651 } \\ & \hline 1.69 \end{aligned}$ |  | CA3015 | 1.401 .15 |  |  |
|  |  | CA3015 |  |  | 1.82 1.50 | CA3060BD |  |
|  |  |  |  | ${ }^{\text {Ca3 }}$ | ${ }_{8.90}{ }^{2} 812182$ | CA3060 |  |
| tor quantity discounts. |  |  |  | CA3018 | 0.880 .72 | CA3060 | 2.742 |
|  |  |  |  | CA3018A | 1.1900 .98 | CA3062 | 2.972 .45 |
| 10 Pln TO-5 I.C. Holders, 20.35 |  |  |  | CA3019 | 0.940 .78 | CA3064 | 1.421 .17 |
|  |  |  |  | A3020 | 1.5651 .28 | CA3065 | 1.421 .17 |
|  |  |  |  | CA3020A | 2.011 .65 | CA3066 | 2.612 .15 |
|  |  |  |  | CA3021 |  | CA3 |  |
|  |  | i.c. Holders, |  |  |  | CA |  |
|  |  |  |  | CA3026 | 1.2551 .03 | CA3071 | 2.011 .65 |
|  |  |  |  | CA 3028 A | 0.88 | CA3073 | 2.07 1.481 .71 1.91 |
|  |  |  |  | CA3029 | 1.1720 .97 0.58 | ${ }_{\text {CA3076 }}$. | ${ }_{1.67} 1.88$ |
| $\begin{array}{llllll}\text { SN7400 } & 0.20 & 0.18 & \text { SN7422 } & 0.48 \\ \text { SN7401 } & 0.20 & 0.48 \\ \text { SN }\end{array}$ |  |  |  | CA3029A | 0.630 .52 | CA3078AT4 | 4.888 .88 |
|  |  |  |  | CA3030 | 1.080 .89 | CA3078' | 2.808 .81 |
|  |  |  |  | CA3030 | 1.641 .35 | CA3079 | 0.940 .78 |
| $\begin{array}{llllll}\text { 8N7403 } & 0.20 & 0.18 & \text { 8N7426 } & 0.32 & 0.2\end{array}$ |  |  |  | CA3033 | 2.922 .41 | CA3080 |  |
| $\begin{array}{lllllll}8 N 7404 & 0.20 & 0.18 & \text { SN7427 } & 0.48 & 0.42\end{array}$ |  |  |  | CA3033 | $4.98{ }^{4.07}$ | CA3080A | 1.901 .57 |
| SN 7405 0.20 0.18 8N7428 0.50 |  |  |  | CA3035 | 1.221 .09 | CA3081 | 1.80 |
| SN7406 $\quad 0.300 .24{ }^{\text {SN7430 }}$ |  |  |  | CA3035 | 1.52 | CA3082 | 1.801 .48 |
| $\begin{array}{llllll}\text { gN7407 } & 0.568 & 0.48 & \text { SN7432 } & 0.48 & 0.42\end{array}$ |  |  |  | CA3036 | 0.78 0.64 | ${ }_{\text {CA3 }}$ | 1.86 <br> 1.71 <br> 1.41 <br> 1.48 |
| SN7408 0.2000 .18 , 8N7433 0.71008 |  |  |  | CA3037 |  | ${ }_{\text {CA }}$ | 1.71 .41 |
| SN7409 $\quad 0.2000018,8 N 7437 \quad 0.520 .45$ |  |  |  | CA3037 | 2.602 .14 | CA3085 | 0.94 0.78 |
| $\begin{array}{lllllll}\text { SN7410 } & 0.20 & 0.18 & \text { 8N7438 } & 0.52 & 0.45\end{array}$ |  |  |  | CA3038 | 2.602 .14 | CA3085A | 1.44.18 |
| $\begin{array}{lllllll}\text { SN } 7411 & 0.22 & 0.20 & \text { SN7439 } & 0.520 .45\end{array}$ |  |  |  | CA3038 | 8.402 .80 | CA3085B | 3.788 .12 |
| $\begin{array}{llllll}\text { 8N7412 } & 0.42 & 0.36 & 8 N 7440 & 0.20\end{array}$ |  |  |  | CA3039 | 1.000 .91 | CA3086 | 0.480 .38 |
| SN7413 0.300027 SN7441AN |  |  |  | CA3040 | 2.958 .48 | CA3088E | 1.701 .51 |
| SN7416 | 0.450 .40 |  |  | CA3041 | 1.301 .07 | CA3089E | 0 |
| 9N7417 | 0.450 .40 | 8N7442 | 790.88 | CA3042 | 1.301 .07 | CA3090Q | 4.70 8.888 |
| $\left\lvert\, \begin{aligned} & \text { 8N7420 } \\ & \text { SN7421 } \end{aligned}\right.$ | 0.200 .18 | 8N7443 | 040.88 | CA3043 | 1.731 .48 | CA3091D | 6.52 6.38 |
|  | 0.220 .20 | 7744 | 04 | CA3044 | 1.421 .17 | E |  |

## LARGEST STOCKS WIDEST RANGE <br> SEMICONDUCTORS \& COMPONENTS <br> BRAND NEW GUARANTEED



## VALVES

SAME DAY SERVICE
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SETS $\begin{gathered}\text { 1R } \delta, 185, ~ 1 T 4, ~ 384, ~ 3 V 4, ~ D A F 91, ~ D F 91, ~ D K 91, ~ D L 92, ~ D L 94 . ~ \\ \text { Set of } 4 \text { ior 21-12. DAF96, DF96, DK96, DL96. } 4 \text { for } \$ 1.55 .\end{gathered}$

| 1R5 |  | 30 Cl 7 | . 78 | EB91 | . 10 | EM80 | - 88 | PCF806 | . 58 | U329 | . 66 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 185 | . 28 | 30 Cl 8 | . 58 | EBC33 | . 40 | EM81 | . 38 | PCF808 | . 68 | U801 | . 80 |
| $1{ }^{14}$ | . 16 | 30 F 5 | . 64 | EBC41 | . 54 | EM84 | -82 | PCL82 | .80 | UABC80 | 88 |
| 384 | . 28 | 30 FLI | . 65 | EBC81 | . 80 | EM87 | - 50 | PCL83 | . 57 | UAF42 | . 50 |
| 3 V 4 | .47 | $30 \mathrm{FL12}$ | -68 | EBC90 | . 28 | EY51 | . 88 | PCL84 | . 34 | UBC41 | . 45 |
| 5U4G | . 81 | $30 F L 14$ | . 88 | EBF80 | . 82 | EY86 | . 29 | PCL85 | .88 | UBF80 | . 84 |
| 6V4G | . 85 | 30 Ll | . 29 | EBF83 | . 89 | EY87 | -29 | PCL86 | . 88 | UBF89 | . 88 |
| 5Y3GT | .84 | 30L15 | . 70 | EBF89 | . 29 | EZ40 | . 48 | PCL88 | . 88 | UCC84 | . 82 |
| 6Z4G | -35 | 30 L 17 | . 67 | ECC81 | . 17 | EZ41 | $\cdot 48$ | PCL800 | . 89 | UCC85 | . 85 |
| 6/30L2 | - 54 | 30P4 | -65 | ECC8: | - 20 | EZ80 | . 28 | PCL805 | $\cdot 88$ | UCF80 | -38 |
| 6AL5 | -11 | 30 P 12 | . 69 | ECC83 | . 85 | EZ81 | . 28 | PENA4 | . 77 | UCH42 | . 58 |
| 6AM6 | -18 | 30 P 19 | -68 | ECC8 | . 84 | EZ90 | -25 | PEN36C | . 70 | UCH81 | . 82 |
| 6AQ5 | -28 | 30 PL 1 | . 60 | ECC804 | - 84 | GZ30 | - 34 | PFL200 | . 62 | UCL82 | . 32 |
| 6AT6 | -20 | $30 \mathrm{PL13}$ | . 89 | ECF80 | . 81 | GZ32 | . 40 | PL36 | . 49 | UCL83 | . 55 |
| 6AU6 | -20 | $30 \mathrm{PL14}$ | -68 | ECF82 | . 26 | KT41 | .77 | PL81 | . 44 | UF41 | . 52 |
| 6BA6 | . 20 | 35L6GT | . 45 | ECH35 | . 55 | KT61 | . 65 | PL81A | . 47 | UF89 | . 30 |
| 6BE6 | . 21 | 35 W 4 | . 25 | ECH 42 | . 59 | KT66 | -78 | PL82 | .31 | UL41 | . 58 |
| 6BJ6 | . 41 | $35 \mathrm{Z4GT}$ | $\cdot 25$ | ECH81 | . 29 | LN 319 | -68 | PL83 | . 88 | UL84 | . 80 |
| 68W7 | - 58 | j0CD6G | -68 | ECH83 | . 40 | LN329 | .72 | PL84. | .80 | UM84 | . 28 |
| 6 F 14 | - 40 | 907 | 49 | ECH84 | . 86 | LN 339 | -65 | PL500 | . 68 | UY41 | -89 |
| $6 \mathrm{~F}^{2} 3$ | -68 | AC/V | . 77 | ECL80 | . 85 | N78 | . 87 | PLū04 | . 68 | UY85 | . 26 |
| $6 \mathrm{~F}^{25}$ | . 68 | B349 | $\cdot 70$ | ECL82 | . 81 | P61 | . 40 | PM84 | . 38 | VP4B | $\cdot 77$ |
| 6 J 7 G | -24 | B729 | . 54 | ECL8B | . 85 | PABC | . 84 | PX25 | . 96 | W77 | . 48 |
| 8K7G | -12 | CCH35 | . 67 | EF39 | . 88 | PC86 | . 47 | PY32 | . 52 | 277 | . 28 |
| 6 K 8 G | $\cdot 17$ | CY31 | . 30 | EF41 | . 60 | PC88 | 47 | PY33 | . 68 | Tranai | ra |
| 697G | . 86 | DAF91 | . 28 | EF80 | . 28 | PC96 | . 42 | PY81 | . 25 | AC107 | -17 |
| 6SL7GT | . 80 | DAF96 | - 88 | EF85 | . 28 | PC97 | - 38 | PY82 | . 25 | AC127 | . 18 |
| 68N7,9T | . 80 | DF91 | -16 | EF86 | 1.30 | PC900 | -29 | PY83 | . 26 | AD140 | .37 |
| 6V6G | . 28 | DF96 | - 38 | EF89 | . 28 | PCC84 | -29 | PY88. | . 38 | AF115 | -20 |
| 6V6GT | . 28 | DH77 | -20 | EF91 | . 18 | PCC8 | . 88 | PY800 | $\cdot 34$ | AF116 | . 20 |
| 6 X 4 | . 28 | DK32 | . 88 | EF9? | . 87 | PCC88 | - 38 | PY80 | -84 | AFliz | . 80 |
| 6X5GT | . 28 | DK91 | . 28 | EF98 | . 85 | PCC89 | . 45 | R19 | .30 | AF 120 | . 17 |
| 10P13 | . 68 | DK99 | . 50 | EF183 | . 28 | PCC189 | . 48 | R20 | $\cdot 70$ | AF127 | . 17 |
| 12AT7 | . 17 | DK96 | . 46 | EF184 | . 81 | PCC805 | . 70 | U25 | . 78 | $\mathrm{OCH}^{26}$ | . 25 |
| 12AU7 | . 20 | DL35 | - 40 | EH90 | . 84 | PCF80 | . 28 | U26 | .70 | OC44 | -18 |
| $12 \mathrm{AX7}$ | -28 | DL92 | . 28 | EL33 | . 65 | PCF80 | . 28 | U47 | . 73 | $0 \mathrm{OC45}$ | -12 |
| 19BG6G | . 80 | DL94 | . 47 | EL34 | . 45 | PCF82 | . 88 | U49 | $\cdot 70$ | 0 C 71 | . 12 |
| 20 F 2 | -67 | DL96 | - 38 | EL4 | . 54 | PCF86 | . 46 | U52 | .31 | OC72 | . 18 |
| 20P3 | $\cdot 75$ | DY86 | . 24 | EL84 | . 23 | PCF800 | . 58 | U78 | -24 | 0c75 | -12 |
| 25LbGGT | - 18 | DY87 | . 24 | -EL90 | . 28 | PCF801 | . 28 | U191 | . 68 | OC81 | . 18 |
| 25 U [GT | . 67 | DY802 | . 88 | EL90 | . 28 | PCF802 | . 28 | U193 | -42 | $\mathrm{OCB1}^{\text {OC8 }}$ | . 18 |
| 30 Cl | $\cdots 28$ | EABC80 | . 82 | EL95 | . 88 | PCF802 | . 58 | U251 | . 68 | $0 \mathrm{OCP}$ | . 18 |
| 30 Cl 5 | . 58 | EAF42 |  | L500 | . 62 | PCF80J | . | U301 | -38 | $\begin{aligned} & \text { OC82 } \\ & 0 C 170 \end{aligned}$ | . 18 |

## 35 TORQUAY GARDENS, REDBRIDGE, ILFORD

 ESSEXMinimum post/packing on 1 valve $7 p$, on each additional valve (3p per valve extra)
Any parcel insured against damage in transit 3 p extra.

## B.H. COMPONENT FACTORS LIMITED

$\mu \mathrm{F}: 0.01,0.015,0.022,0.033,0.047,3 \mathrm{p}$ each; $0.068,0.1,0.15,4 \mathrm{p}$ each; $0.22,5 \mathrm{p}$
 160V: ( $\mu \mathrm{F}$ ) $0.01,0.015,0.022,2 \mathrm{p} ; 0.047,0.068,3 p ; 0.15,0.22,4 p ; 0.33,5 p$; 400 V : ( $\mu \mathrm{F}) 0.001,0.0015,0.0022,0.0033,0.01,2 p ; 0.015,0.033,3 p ; 0.068,4 p$ MINIATURE ELECTROLYTIC MULLARD C426 SERIES (5p each) ( $\mu$ F/V) 0.64/64, 1.6/20, 4/4, 8/40, 10/40, 10/64, 16/40, 2064, 25/25, 32/10, 40/16, 64/10, 80/16, 80/25, $100 / 6 \cdot 4,125 / 16,200 / 6 \cdot 4,200 / 10,320 / 6 \cdot 4,125 / 10$. MULLARD C437: ( $\mu \mathrm{F} / \mathrm{V}$ ) 64/64, 9p; 160/25, 9p; 160/40, $11 \mathrm{p} ; 640 / 6 \cdot 4,9 \mathrm{p}$; $1600 / 6 \cdot 4,14 \mathrm{p}$.

ELECTROLYTIC CAPACITORS. Tubular and large can ( $\mu$ F/V) $2.5 / 50,3$; ; $4 / 10,10 / 25,16 / 15,20 / 25,25 / 15,25 / 25,40 / 6,64 / 10,200 / 6$ $250 / 10$, 4 p; 10/6, 10/50, 25/50, 32/50, 50/10, 64/25, 100/25, 5p; 50/50, 64/40, $250 / 15,1000 / 3,6 \mathrm{p}$ : $100 / 50,250 / 25,400 / 10,500 / 10,500 / 12,640 / 10,1,000 / 6$ 8p; 500/25, 10p; 500/50, 12p; 1,000/12, 10p; 1,000/25, 2,000/12, 2,500/12, 15p 1,000/50, 35p ; 2,000/25, 25p; 2,500/25, 30p; 2500/50, 55p; 3.000/50, 65p 5,000/50, 85p.

CERAMIC PLATE CAPACITORS
$750 \mathrm{~V}:(\mathrm{PF}) 5,10,25,40,70,220,2 \frac{1 \mathrm{p}, ~(\mu F / V) 0.0047 / 30,0.01 / 350,2 p}{}$ $0.047 / 30,3 \mathrm{p} ; 0.1 / 30,4 \mathrm{p} ; 0.1 / \mathrm{i} 00,5 \mathrm{p} ; 22 \mathrm{pF}-1000 \mathrm{pF} 50 \mathrm{~V}$, E12 Series; 1500 pF ;-
$0.022 \mu \mathrm{~F} 50 \mathrm{~V}$ E6 Series 2 p $0.022 \mu \mathrm{~F} 50 \mathrm{~V}$, E6 Series 2 p each.
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A PROFESSIONAL $\xlongequal{50 \overline{\mathrm{\mu A}}}$
sk $H$
BRAND NEW AT A FRACTION OF THE ORIGINAL PRICE! 3 digit plus polarity indicator. 1 Its 17 positions allow it to measure in the
following ranges: d.c. volts 100 V to 999 V a ac, volts $100, \mathrm{~V}$ to 420 V ; d.c.
 $100 \mu \mathrm{~A}$; a.c. $100 \mu \mathrm{~A}$. Resistance $0.1 \Omega$ to 999 k a. Operates from 230 V 50 Hz $12 v$ d.c.external.


## BRAND NEW DIGITAL PANEL VOLTMETERS 10 mV to $\mathrm{Z}=1.99 \mathrm{~V}$, 199 measuring points. Input An $Z=100 \mathrm{Mn}$ Automatic zeroing. Dimensions $\underset{\text { Orizinal }}{72 \mathrm{Mm}} \mathrm{H} \times 72 \mathrm{~mm}$ Price $\mathrm{ES2}$. <br> P. F. RALFE <br> IO CHAPEL STREET, LONDON, N.W.I Tel. 01-723 8763

## THIS IS THE FIRST PAGE OF THE GREAT BI-PAK SECTION

## BRAND NEW FULLY GUARANTEED DEVICES

$\begin{array}{llll}\mathrm{ACl07} & 0.20 & \text { ADI } 62 & 0.38\end{array}$ $\begin{array}{lll}A C 113 & 0.20 & A \\ A C 115 & 0.28 & A \\ A C 117 K & 0.20 & \\ A C 122 & 0.12 & A \\ A C 125 & 0.17 & A \\ A C 126 & 0.17 & A\end{array}$
$\mathrm{AC126}$
ACl 27
$\mathrm{AC128}$
$\mathrm{AC132}$
$\mathrm{AC132}$
$\mathrm{AC134}$
$\begin{array}{ll}\mathrm{AC137} & 0.14 \\ & 0.14\end{array}$
4 Cl 141
$\mathrm{AC14}$

${ }_{4 C 151}$
AC155
$\mathrm{AC156}$
$\mathrm{AC157}$
AC165
ACl65
ACl 66
ACl 66
AC 167

| ACl 67 |
| :--- |
| $\mathrm{AC1}$ |
| AC |

AC168
AC169
AC176
AC176
AC177
ACl 177
ACl 8
ACl
$\mathrm{AC179}$
$\mathrm{AC180}$
$\mathrm{AC180K}$
$\mathrm{AC181}$
AC181
$A$ C181K
AC187
AC187K
${ }_{4} \mathrm{Cl} 188 \mathrm{~K}$
ACl 88
ACl 8 B
$A C Y$
ACY
$A C Y$
$A C Y$
$A C P$


| CY |
| :---: |
| CY | CY $A C Y$

$A C Y$
$A C Y$
CY 46

D130
AD143

 o000000000000000 00000000000000000000000000000000000000000000

2G371
2 G 371
$2 \mathrm{Ca71}$

2 N 2219 $2 \mathrm{~N}^{2} 2120$
0.16
0.12
0.17


0.20
0.22


#### Abstract

2 N 3054 2 N 3055


0.46
0.50
0.14
0.16
0.14
0.14
0.14
0.17
0.21
0.21
0.28
0.42
0.15
0.15
0.28
0.28
0.75
0.09
0.10
0.10
0.21
0.10
0.09
0.11
0.07
0.09
0.09
0.09
0.28
0.50
0.35
0.28
0.28
0.30
0.28
0.27
0.12

2N 4059
$2 N 4060$ 2N4060
NN4061
2N 4062 2 N 4061
$2 \mathrm{~N} 40 \mathrm{~B}_{2}$ 2N4284
2 N 4285
2 N 4286 2 N 4287 2N 4288 2 N 4289 2N4290 2 N 4292 $2 N 4293$
$2 N 5172$ 2N5457 2 N 545
28301 28302 ${ }_{28302}$ 28303 28305
28306 28306
28307 28307
28321 28322
28322 A 28323
2832
28325 28325
28327 3701 40362 0.10
0.12
0.12
0.12
0.17
0.17
0.17
0.17
0.17
0.17
0.17
0.17
0.17
0.17
0.12
0.32
0.32
0.40
0.80
0.42
0.42
0.55
0.70
0.84
0.84
0.84
0.58
0.42
0.42
0.58
0.70
0.70
0.70
0.70
0.48
0

DIODES AND RECTIFIER
 0.35
0.07
0.07
0.07
0.07
0.09
0.06
0.08
0.07
0.08
0.07
0.05
0.05
0.07
0.07
0.06
0.06
0.06

| AA119 | 0.08 | BY133 | $0 \cdot 21$ | OAl0 | 0.35 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AA120 | 0.08 | BY164 | 0-50 | $0 \mathrm{A47}$ | 0.07 |
| Af129 | 0.08 | BYX38/30 |  | OA70 | 0.07 |
| AAY30 | 0.09 |  | $0 \cdot 42$ | 0.479 | 0.07 |
| AAZ13 | 0.10 | BYZ10 | 0.35 | OA81 | 0.07 |
| BA100 | 0.10 | BYZ11 | $0 \cdot 30$ | OA85 | 0.09 |
| BA116 | 0.21 | BYZ12 | $0 \cdot 30$ | OA90 | 0.06 |
| BAl26 | 0.22 | 13YZ13 | $0 \cdot 25$ | OA91 | 008 |
| BA148 | 0.14 | 13YZ16 | 0.40 | OA95 | 0.07 |
| BA154 | 0.12 | BYZ17 | 0.35 | OA200 | 0.08 |
| BA155 | 0.14 | BYZ18 | $0 \cdot 35$ | OA202 | 0.07 |
| BA156 | $0 \cdot 13$ | HYZ19 | $0 \cdot 28$ | SD10 | 0.05 |
| HY100 | 0.15 | CO62 |  | 8 SI 19 | 0.05 |
| BY101 | 0.12 | (Eg) OA91 |  | IN34 | 0.07 |
| B) 105 | $0 \cdot 17$ |  | 0.05 | 1 N34A | 0.07 |
| BY114 | 0.12 | CG651 |  | 1 N 914 | 0.06 |
| BY126 | 0.14 | (Eq) OA70. |  | 1N916 | 0.06 |
| BY127 | 0.15 | OA79 | 0.06 | IN414B | 0.06 |
| BY128 | $0 \cdot 15$ | OA5 | $0 \cdot 35$ | 18021 | 0.10 |
| BY130 | 0.16 | OA5sL | 0.21 | IN951 |  |

## NEW COMPONENT PAK BARGAINS

No
C 1250 Resiators mized values approx. coust by weight
200 Capacitors mixed values approx. count by weigh

```
th W Resistors mixed preferred values
```

Pieces assorted Ferrite Rul
Tuning Gangs, MW/LW VHF
Pack Wire 50 metres assorted colours
Meer Switches
Assorted Pots \& Pre-Sets
Jack sockets $3 \times 3.5 \mathrm{~m} 2 \times$ Standard 8 witch Types
$\begin{aligned} 5 & \text { Jack Sockets } 3 \times 3.5 \mathrm{~m} 2 \times \text { Standard } 8 \text { witch T } \\ 40 & \text { Paper Condensers preferred trpes mixed values }\end{aligned}$
1320 Electrolytice Trans. types
C14 1 Pack assorted Hardware-Nuts/Bolte, Grommets elc. Mains Toggle 8witches, 2 Amp D/P
Assorted Tag Strips \& Panela
Assorted Control K nolys
Rotary Wave Change Switches
193 Relays 6-24V Operating
204 Sheets Copper Laminate approx. $10^{\circ} \times 7$
a pack Nos. C1, Ca Cis co

## PLUS-MUCH MORESEND NOW FOR THE

BI-PAK "Component Catalogue"
$5 p$ to cover postage etc

THE NEW S.G.S. EA 1000 AUDIO AMP MODULE *Guarantee

## 3 Watts

R.M.S.

ONLY
£2.63 each
Modual Tested and Guaranteed
Qty. 1-9 $28.63: 10-2542 \cdot 28$ Price each. Larger quantities quoted on request. Full hook-up diagrams modual or avallable separately at 10 p eac


## SYSTEM 12 STEREO

Each Kit contains two Amplifier Modules, 3 watts RMS, two loudspeakers, 15 ohms, the pre-amplifier, transformer, power supply module, front panel and other accessories, as well as an illustrated stage-by-stage instruction booklet designed for the ONLY
 Further details available on $54=15$ request.

## JUMBO COMPONENT PAES MJXEDELECTHONTC COMPONENTS Exceptionally good value

Resistors, capacitors. pots, electrolytics Approximately 31bs in weight. Price incl. P. \& P. \&1. 50 only.

## $\left(\begin{array}{l}y^{\circ} \\ 1 \\ 2\end{array}\right.$ <br> BRAND NEW POST OFFICE TYPE TELEPHONE DIALS ONLY 75p each

$\qquad$

## The largest selection

NEW LOW PRICE TESTED S.C.R.'s
 $\begin{array}{rlllllll}50 & 0.23 & 0.25 & 0.35 & 0.35 & 0.47 & 0.50 & 0.53 \\ 100 & 0.25 & 0.33 & 0.47 & 0.47 & 0.850 & 0.58 & 0.63 \\ 100\end{array}$ $\begin{array}{llllllllllll}200 & 0 & 35 & 0.37 & 0.49 & 0.49 & 0 & 57 & 0.61 & 0.75 & 1.60\end{array}$ $\begin{array}{lllllllll}400 & 0.43 & 0.47 & 0.56 & 0.56 & 0.67 & 0.75 & 0.93 & 1.75\end{array}$ $\begin{array}{llllllllll}600 & 0.83 & 0.57 & 0.68 & 0.68 & 0.77 & 0.97 & 1.25 & - \\ 800 & 0.63 & 0.70 & 0.80 & 0.80 & 0.90 & 1 \cdot 20 & 1 \cdot 50 & 4.00\end{array}$

## SIL. RECTS. TESTED

| PIV 300 mA | 750 mA |
| :---: | :---: |
| 500.04 | 0.05 |
| 1000.04 | 0.06 |
| 2000.05 | 0.09 |
| 4000.06 | 0.13 |
| 60000.07 | 0.16 |
| 8000.10 | 0.17 |
| 1000 | 0.11 |
| 1200 | 0.25 |

triacs
VBOM 2A 6A 10A
TO-1 TO-66 TO. 88
$\begin{array}{ll}100 & 80 \\ 200 & 80 \\ 400 & 70\end{array}$

## DIACs

FOR
TRIACS
BR100 (D32) 87p each
FREE
On ${ }^{80 p}$ Puk of your
own chotce free with own cholce free with
orders ralued 24 or over.
BRAND FEW TEXAB GERM. TRANSIETORS Coded and Guaranteed

## Pak No.



2N2060 NPN 8IL. DUAL TRARS. CODE D1699 TEXA8. Our price 25p nech.

120 VCB MIXIE DRIVER TRANSISTOR. 8 Sim . BSX21 * C407, 2N1893 CODED ND 120 . $1-24$ 7 D each. TO.

## ail. trans. suitable for P.E. Organ. Metal TO-18

## POWER <br> TRANS

BONANZA!

KING OF THE PAKS Unequalled Value and Quality SUPER PAKS NEW BI-PAK UNTESTED

QUALITY TESTED bemicondoctors

|  |  | $\underset{\text { Price }}{\text { Price }}$ |
| :---: | :---: | :---: |
| Q1 | 20 Red spot transistors P | $0 \cdot 50$ |
| Q2 | 16 White rtot R.F. transistors PNP | 0.80 |
| Q3 | 4 OC 77 type transistors | $0 \cdot 30$ |
| Q4 | 6 Matched transistors OC4 | 0.50 |
| Q5 | 4 OC 75 trankistors | 80 |
| Q6 | 5 OC 72 transistors | 0 |
| Q7 | 4 AC 128 transistors PNP | 50 |
| Q8 | 4 AC 126 transistors PNP | 50 |
| Q9 | 7 OC 81 type transistors |  |
| Q10 | 7 OC 71 type transigtors |  |
| Q11 | 2 AC 127/128 Complementary pairs PNP/NPN |  |
| Q12 | 3 AF 116 type transistory |  |
| Q13 | 3 AF' 117 trpe transiator3 |  |
| Q14 | 3 OC 171 H.F. trpe transintors |  |
| Qi5 | 7 2N2926 sil. Epoxy transistors mixed colours .. |  |
| Q16 | 2 GET880 low noise Germsnium transistora |  |
| Q17 | 5 NPN $2 \times 8 \mathrm{~T} .141 \& 3 \times$ ST. 140 |  |
| Q18 | 4 MADT'S $2 \times$ MAT $100 \& 2 \times$ MAT | $0 \cdot 50$ |
| Q19 | 3 MADT'S $2 \times$ MAT $101 \& 1 \times$ MAT 121 |  |
| Q20 | 4 OC 44 Germanium transiat ors A. |  |
| Q21 | 4 AC 127 NPN Germanium transigtorn |  |
| Q22 | 20 NKT transistors A.F. F.F. colled |  |
| Q23 | 10 OA 202 Silicon diodes sub |  |
| Q24 | 8 OA 81 diodes |  |
| Q25 | 15 IN914 Silicon diodes 75 PIV 75 mA |  |
|  | 8 OA9s Germanium dioles aub-min IN 69 |  |
| Q27 | 2 10A PIV Silicon rectifiers 18425 | 30 |
| Q28 | 2 8ilicon power rectifiers BYZ 13 |  |
| Q29 | $\begin{aligned} & 4 \text { Bilicon transistors } 2 \times 2 N 696 \text {. } \\ & 1 \times 2 \mathrm{~N} 697,1 \times 2 \mathrm{~N} 698 \end{aligned}$ | 0.50 |
| Q30 | 7 Silicon switch transistors 2N706 | 0.50 |
| Q31 | 6 8illeon switch transistors 2 N 708 | 0.50 |
| 2 | ```3 PNP gilicon transistors 2 < 2N113 1\times2N1132``` |  |
| Q33 | 3 Sillcon NPN transistora 2N1711. |  |
| Q34 | 7 silicon NPN transistors 2 N2369, 500 MHz (code P397) | $0.80$ |
| Q35 | $\begin{aligned} & 3 \text { 8ilicon PNP TO-5. } 2 \times 2 N 2904 \text { \& } \\ & 1 \times 2 N 290 \pi \end{aligned}$ | 0.80 |
| Q35 | $72 \mathrm{~N} 3646 \mathrm{TO}-18$ platic 300 MHz NPN |  |
| Q37 | 3 2N3053 NPN 8licon transistorg | 0.50 |
| Q38 | 7 NPN transistors $4 \times 2 N 3703,3 \times$ 2N 3702 | $0.50$ |

## ELECTRONIC SLIDE-RULE

The MK side Rule, designed to simplify Elec tronic calculations festures the following scales:-
Conversion of Frequency and Wavelength Calculation of $L$, $\cdot$ and fo of Tuned Circuits Reartance and self Inductance. Area of Circles. Volume of Cyifinders. Resistance of Conductors. Weight of Conductors. Decibel Calculations. Angle Functions. Natural Logs end ' $e$ ' Functions. Multipllcation and Division. Equaring, Cublng and 8 quare Roots. Conversion of kW and Hp . A must for every electronic engineer and enthusiinstructions. $\quad$ Prire each: $23 \cdot 35$

## GENERAL PURPOSE GERM. PNP

Coded GERERAL PURPOSE GERM. PNP
 $2 N 456 A-457 A-458 A, 2 N 511$ A \& B. 2G220-222, ETC.
VCBO 80 V VCEO 50 V IC 10A PT. 30 WATTE Hfe VCBO 80 V VCEO 50 V IC 10A PT. 30 WATTS Efe
$30-170$. $30-170$.
PRICE

$$
4{ }_{48}^{1-24}
$$

$\stackrel{25-99}{40 \mathrm{p} \text { each }}$ 100 up sTLICON High Voitage 250V HPN TO-3 case. G.P. Bwitchligg \& Amplifier
Applicat lons. Brand new Coded $\mathbf{R} 2400$ VCBO 250/VCEO 100/IC 6A/30 Watte. GFE type 20/IT 5MHZ.
UR PRICE FACH


## 2N3055

115 WATT 8IL POWER KM
50 DEACH

FULL RANGE OF VOLTAGE RANGE 2-39V. 400 mV (DO.7 Case) 18p ea. 1\% (TopHat) $18 p$ ea. 10 w ( $80-10$
8 tud) 25 p es. 8tud) 25 en. All fully
tested $5 \%$ tol. and tested
marke marked. State voltage
required. 10 amp POTTED
BRIDGE RECTIFIER on heat sink. 100PIV. 90p each

## NEW LINE

Plastic Encapanlated 2Amp. BRIDGE RECTS. 50 ₹ RMS 82p each 100 V RMS
400 R RMS Size $15 \mathrm{~mm} \times 6 \mathrm{~mm}$

## UEIJUNCTION

UT48. Eqvt. $2 N 2646$,
Eqvi. TIS43. BEN3000 27p each, 25-99 25p

CADMIUM CELLS
ORP12 48p
ORP60, ORP61 40peach
GEMERAL PURPOSE IPN SILICON BWITCEING TRANS. TO- 18 27/R8/9sA. Ali usable circults. ABLE in PNP Bim. to 2N2906, BCY70. When
ordering pleque


#### Abstract

ordering preterence


|  |  | Fip |
| ---: | ---: | ---: |
| 20 | For | 0.50 |
| 50 | For | 1.00 |
| 100 | For | 1.75 |
| 500 | For | 7.50 |
| 1000 | For | 18.00 |

SIL. G.P. DIODES Ip $\begin{array}{lr}300 \mathrm{~mW} & 30.0 .50 \\ 40 \mathrm{PIV}(\text { Mln ) } & 100.150\end{array}$ Sub-Min. 800 . 5.00 Full Tested $1,000.900$ Full Tested 1,000, 9.00

## ADI61/162

M/P COMPGERM TRANS.
OUR LOWEST PRICE OF
S5p PER PAIR

## SILICOH 50 WATT8 EATCHED MPN/PNP

BIP 19 NPN TO-3 Plastic. BIP 20 PNP. Brand new VCBO 100/VCEO 50/IC 10A. HFE type 100/ft 3nIH? OUR PRICE PER PAIR

25-99 pre. 65p $\quad 100$ pre. 50p

## NEW EDITION 197

TRAESISTOR EQUIVALENTS BOOK. A comolete cross reference and equivalents book for European, American and Japanese Transid-
tors. Exclusive to BI-PAK 00p Anleric
tors.
each.

A LARGE RANGE OF TECHNICAL AND DATA BOOKS ARE NOW AYAILABLE EX. 8TOCK. 8END FOR FREE LIST.

> SILICON PHOTO TEAK AISTOR, TO-18 Lens en APN BIm. To $P$ PW Full dat vallable Fully guaranteed Qty. 1.2425 .90100 up
> Price each 460 40p 85p

## F.E.T.'S

## 2N3819 85 2 N 5458

2N3820 80D 2N5459
2N3821
2N 3823

Iftegrated circuit paigs
Manufanturers ' Fall Onta' which incluile Functional and Part-Functional Unita These are classed as 'out-of-spec' from the maker's very rigid epecffestions, but are ldeal for learning about I.C's and experimental work.

Pak No. Contente Price $\begin{array}{ll}\text { UIC0 }=12 \times 7400 & 0.80 \\ \text { UIC0 }=12 \times 7401 & 0.50 \\ \text { UIC0 }\end{array}$ $\begin{array}{ll}\text { UICO }=12 \times 7401 & 0.60 \\ \text { UIC0 } 2=12 \times 7402 & 0.50\end{array}$ $\begin{array}{ll}\text { UICO3 }=12 \times 7403 & 0.50 \\ \text { UICO }=12 \times 7404 & 0.50\end{array}$ $\begin{array}{ll}\mathrm{UICOS}=12 \times 7405 & 0.50 \\ \mathrm{U} 1 \mathrm{CO}=8=8 \times 7406 & 0.50\end{array}$ $\begin{array}{ll}\text { UIC07 }=8 \times 7407 & 0.80 \\ \text { UIC10 }=12 \times 7410 & 0.60\end{array}$ | UIC1 $3=8 \times 7413$ | 0.60 |
| :--- | :--- |
|  | 0.50 |
|  | $=1(2)=12 \times 7420$ | $\begin{array}{ll}\text { UIC2 } & =12 \times 7420 \\ \text { UIC30 } & 0.60 \\ \text { UIC4 } & =7430 \\ 0.50 \\ \text { U }\end{array}$ $\begin{array}{ll}\text { UIC40 }=12 \times 7440 & 0.50 \\ \text { UIC4 } 1=5 \times 7441 & 0.50\end{array}$ $\begin{array}{ll}\mathrm{UIC} 42=5 \times 7442 & 0.50\end{array}$ $\mathrm{U1C4} 3=5 \times 7443 \quad 0.50$ $\begin{array}{ll}\mathrm{VIC45}=5 \times 7444 & 0.50 \\ 0.50\end{array}$ BI-PAES AEW COMPONERE SHOP HOW OPEF WITH A WIDE RAKGE OF ELECTRONIC COMPONENTS AND ACCESSORIES AT COMPETITIVE PRICES I8 BALDOCK STREET (AIO), WARE, HERTS. TEL. (STD 0920) 61593.

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All mail orders please add $10 p$ post and packing. Send all orders to B/-P.AK P.O. Box 6 ,
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74 Series T.T.L. I.C's
BI-PAE ETLLL LOWEST IN PRICE. FULL SPECIFICATION


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Price each
Eрогу TO-6 саее 1-24 28-99 100 up口L900 Buffer 85p 38p 87p oL914 Dral $21 / p$ 8-1 85 28p 27p口L923 J-K fip-fop 50p 47p 45p Date and Circaita Booklet for IC'a
Price 7p.

DUAL IT LWIE BOCEETS.

$$
\begin{aligned}
& 14 \text { \& } 16 \text { Leed sockets for weo with } \\
& \text { DUAL-IN-LINE I.C'a. TWO Rangea } \\
& \text { PROFEGSIONAL A NEW LOW COBT. } \\
& \text { PROF TYPE NO. NRW LOW COBT } \\
& \begin{array}{llll}
\text { T8O } 14 \text { plntype } & 30 \mathrm{p} & 27 \mathrm{p} & 25 \mathrm{p}
\end{array}
\end{aligned}
$$

$$
\begin{aligned}
& \text { LOW COBT No. } \\
& \begin{array}{l}
\text { BPg } 14 \\
\text { BPg } 16
\end{array}
\end{aligned}
$$

# BI-PAK DO IT AGAIN! 50W pk 25w (RMS) 

$0.1 \%$ DISTORTION HI-FI AUDIO AMPLIFIER
$\star$ Frequency Response 15 Hz to ONLY 100,000-1dB.

* Load-3, 4, 8 or 16 ohms.
$\star$ Distortion-better than $\cdot \mathbf{1 \%}$ at 1 KHz .
©3.251 each
* Supply voltage $10-35$ .
$\star$ Overall size $63 \mathrm{~mm} \times$ $105 \mathrm{~mm} \times 13 \mathrm{~mm}$.
* Signal to noise ratio 80 dB .

Volts.

Failor made to the most atringent specilication using top quality components and incorporating the latest sold state circuitry and ALSO was concelveri to All he need for all your A.F. amplification needs.

## STABILISED POWER MODULE SPM80

AP80 is eapecially deaigned to power 2 of the AL50 Ampliflers, up to 15 watt (r.m.s.) per channel almul and clrcuit techniques incorporating complete short and cormer MT80, the unlt will provide outputs of up to 1.5 amps at 35 volts. Bize: $63 \mathrm{~mm} \times 105 \mathrm{~mm} \times 30 \mathrm{~mm}$.
These units enable you to build Audlo Syatems of the highest quality at a hitherto unobtainable price. Also idesl for many
other applications including:-Diso Systems, Public Address, Intercom Units, etc. Handbook available, 10p PRICE £2.95
TRANSFORMER BMT80 £1 95 p. \& p. 25p.

## STEREO PRE-AMPLIFIER TYPE PA100

Bullt to a specification and NOT a price, and yet still the grestest value on the market, tae PAlo0 stereo pre-amplifier has been conceived from the latest circuit tecnniques o less than eight silicon Ahe more NPN devices for use in the input stage
Three switched stereo inputs, and rumble and acratch filtera are features of the PAl00, which also has a sTEREO/MONO switch, volume, balance and continuousiy variable base and treble controls.

SPECIFICATTON
Frequency Response Harmonic Distortion Inputs: 1. Tape Head better than $0.1 \%$
2. Radio, Tuner $\quad 35 \mathrm{mV}$ into $50 \mathrm{~K} \Omega$
3. Magnetic P.U. $\quad 1.5 \mathrm{mV}$ into $50 \mathrm{~K} \Omega$

All input voltages are for an output of 250 mV . Tape and P.U. Inputa equalised to RIAA curve withln $\pm 1 d B$. from 20 Hz to 20 KHz . Bass Control
$\pm 15 \mathrm{~dB}$ at 20 Hz
$\pm 15 \mathrm{~dB}$ at 20 KHz
8 KHz
better than - 65 dB
+26 dB
+35 vol
+35 volts at 20 mA
$292 \mathrm{~mm} \times 82 \mathrm{~mm} \times 36 \mathrm{~mm}$
ONLY £11-95
SPECIAL COMPLETE KIT COMPRISING 2 AL50's, 1 SPM80, 1 BMT80 \& 1 PA100 ONLY $233 \cdot 00$ FREE p. \& p.
All prices quoted in new pence Giro No. 388 - 7006
Please send all orders direct to warehouse and despatch department

# YATES ELECTRONICS (FLITWICK) LTD. <br> ELSTOW STORAGE DEPOT KEMPSTON HARDWICK BEDFORD 

C.W.O. PLEASE, POST AND PACKING PLEASE ADD IOp TO ORDERS UNDER $k 2$.

Catalogue which contains data sheets for most of the components listed will be sent free on request. 10p stamp appreciated.

OPEN ALL DAY SATURDAYS

RESISTORS
W lakra high seability carbon film-very low noise-capless construction IW Mullard CR25 carbon film-very small body size $7.5 \times 2.5 \mathrm{~mm}$ 2\% ELECTROSIL TRS

| Power |  |  | Values | Pri |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| watts | Tolerance | Ranze | available | 1-99 | $100+$ |
| $\pm$ | 5\% | 4.7n-2.2Mn | E24 | 1 P | ${ }^{0.8 p}$ |
| , | 10\% | $3.3 \mathrm{Mn}-10 \mathrm{Mn}$ | E12 | 1 p | ${ }^{0.8 p}$ |
|  |  | 10n-1Mn | E24 | $3.5 p$ | 3p |
| + | 10\% | 4 l | E12 | Ip | 0.tp |
| 4 | 10\% | $4.7 n-1 m n$ $18-10 n$ | E12 | 1p |  |
| Quant |  |  | frac |  |  |

## DEVELOPMENT PACK

0.5 watt $5 \%$ lskra resistors 5 off each value 4.70 to 1 Ma

E12 pack 325 resistors 62.40. E24 pack 650 resistors $\mathbf{4 4} 7 \mathbf{7 0}^{\circ}$

## POTENTIOMETERS

Carbon track 5 kn zo 2 Mn , los or linear (los $\ddagger \mathrm{W}$, lin $\ddagger \mathrm{W}$ ) Single, 12p. Dual gang (stereo), 40p. Single D.P. switch 24p.

## SAELETON PRESET POTENTIOMETERS

Linear: $100,250,5000$ and decades to 5 Mn . Horizontal or vertical P.C mounting ( 0.1 matrix)
Sub-miniature $0.1 \mathrm{~W}, 5_{p}$ each. Miniature 0.25 W , $6 p$ each

## TRANSISTORS

| ACIOT | 15p | BCIO8 | 10p | 8FY51 | 22 p | OCP7I | 40p | 2N37 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ACI26 | 12p | BC109 | 10p | BFY52 | 22p | ORPI2 | 50p | 2N370 | $13 p$ |
| AC127 | 12p | BC147 | 10p | BSY 56 | 32p | 2N2369 | 16p | 2N3705 | 12p |
| AC128 | 12p | BC148 | 13p | OC26 | 45p | 2N2646 | 60p | 2N370 | $11 p$ |
| ACI3 | 12p | BC149 | 13p | OC28 | 45p | 2N2926 | 9p | 2N370 | 12p |
| AC132 | 12p | BC157 | 13p | OC35 | 45p | 2N2926 | 9p | 2N3709 | 10p |
| ADI40 | 50p | BC158 | 13p | $\bigcirc \mathrm{OC} 42$ | 12p | 2N2926 | $\mathrm{OP}^{\text {P }}$ | 2N3709 | $11 p$ |
| AD161 | 33p | BC159 | 13p | OC44 | 12p | 2N2926 |  | 2 N 37 | 11p |
| AD162 | 36p | 80131 | 75p | 0 C 45 | 12p |  | 10p | 2 N 37 | 11p |
| AFll4 | 20p | BD132 | 75p | OC70 | 12p | 2N3054 | 58p | 2 N 406 | 12p |
| AFIIS | 20p | BFI79 | 32p | OC71 | 12p | 2 N 3055 |  | ZTX | 15p |
| AFII6 | 20p | BF181 | 25p | OC72 | 12 p | 2N3055 |  | $\geq \mathrm{T} \times 5$ |  |
| AFII7 | 20p | BF194 | 15p | OC72 |  | 2N3442 |  | $Z T \times 5$ $7 \times 50$ |  |
| AFI'8 | 38p | BF195 | 15p | $0 \mathrm{OC81}$ | 12p |  | 140p | ZTX5 | 16p |
| BCl07 | 10p | BFY50 | 22p | OC82D | 12p | 2N3702 | 13p | 40362 | 58p |
| ZENER DIODES $400 \mathrm{~mW} 5 \%$ 3.3V to $30 \mathrm{~V}, 15 \mathrm{p}$. |  |  |  | LINEAR I.C.'s(D.IL.)$70950 \mathrm{p} 741.50 p$$71050 p 748,50 p$ |  |  |  | DIL Socket 14 and 16 pin. 16p |  |
|  |  |  |  |  |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  |
| DIODES <br> RECTIFIER <br> SIGNAL |  |  |  |  |  |  |  |  |  |
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| BY127 | $1250 V$$800 V$ |  | IA | $12 p$ |  | OAB5 7p |  |  |  |
| BZV10 |  |  | 6 A | $25 p$ |  | OA90 5p |  |  |  |
| BZY13 | 200 V |  | 6 A |  |  |  |  |  |  |  |  |
| IN4001 | 50 V |  | IA | 7p |  | OA202 7p |  |  |  |
| IN4004 | 400 V |  | 1 A | 12p |  | IN4148 5p |  |  |  |
| IN 4007 |  |  | IA |  |  |  | BA |  | 8p |

ERUSHED ALUMINIUM PANELS
$12 \mathrm{in} \times 6 \mathrm{in}=25 \mathrm{p} ; \quad 12 \mathrm{in} \times 2 \mathrm{tin}=10 \mathrm{p} ; \quad$ in $\times 2 \mathrm{in}=7 \mathrm{p}$
SLIDER POTENTIOMETERS
$86 \mathrm{~mm} \times 9 \mathrm{~mm} \times 16 \mathrm{~mm}$, length of track 59 mm
SINGLE 10k $\Omega, 25 \mathrm{k} \Omega, 50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega$, log or 1 in
DUAL GANG 10k $n+10 k n$, ete, loz or lin
Knob for above
20 gauge panel 12 in $\times 4$ in with slots cut for use with slider pots
Grey or matt black finish, complete with fixings for 4 pots.

MULLARD POLYESTER CAPACITORS C296 SERIES
$400 \mathrm{~V}: 0.001 \mu \mathrm{~F}, 0.0015 \mu \mathrm{~F}, 0.0022 \mu \mathrm{~F}, 0.0033 \mu \mathrm{~F}, 0.0047 \mu \mathrm{~F}, 21 \mathrm{p}, 0.0068 \mu \mathrm{~F}, 0.01 \mu \mathrm{~F}$ $0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 0.033 \mu \mathrm{~F}, 3 \mathrm{p} .0 .047 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 0.1 \mu \mathrm{~F}, 4 \mathrm{p}, 0.15 \mu \mathrm{~F}, 6 \mathrm{p} .0 .22 \mu \mathrm{~F}, 71 \mathrm{p}$. $0.3 \mu \mathrm{~F}, 01 \mathrm{p}, 0.47 \mu \mathrm{~F}, 1 \mathrm{P}$,
$160 \mathrm{~V}: 0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 0.033 \mu \mathrm{~F}, 0.047 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 3 \mathrm{p}, 0.1 \mu \mathrm{~F} 31 \mathrm{p} .0 .15 \mu \mathrm{~F} 4 \mathrm{p}$. $0.22 \mu \mathrm{~F}$, 5p. $0.33 \mu \mathrm{~F}, 6 \mathrm{p} .0 .47 \mu \mathrm{~F}$, 7 1p. $0.68 \mu \mathrm{~F}$, $11 \mathrm{p} .1 .0 \mu \mathrm{~F}$, 13 p .
MULLARD POLYESTER CAPACITORS C280 SERIES
250 V P.C. mounsing: $0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 3 \mathrm{p}$. $0.033 \mu \mathrm{~F}, 0.047 \mu \mathrm{~F}, 0.060 \mu \mathrm{~F}$,


MYLAR FILM CAPACITORS $100 V$ $0.001 \mu \mathrm{~F}, 0.002 \mu \mathrm{~F}, 0.005 \mu \mathrm{~F}, 0.01 \mu \mathrm{~F}, 0.02 \mu \mathrm{~F}$ 21p. $0.04 \mu \mathrm{~F}, 0.05_{\mu} \mathrm{F}, 0.068 \mu \mathrm{~F}, 0.1 \mu \mathrm{~F}, 31 \mathrm{p}$.

CERAMIC DISC CAPACITORS 00pF to $10,000 \mathrm{pF}, 2 \mathrm{peach}$

## ELECTROLYTIC CAPACITORS—MULLARD C426 SERIES

$(\mu F / V) 10 / 2 \cdot 5,40 / 2 \cdot 5,80 / 2 \cdot 5,160 / 2 \cdot 5,320 / 2 \cdot 5,500 / 2 \cdot 5,8 / 4,32 / 4,64 / 4,125 / 4,250 / 4$. $400 / 4,6.4 / 6.4,25 / 64,50 / 6.4,100 / 6 \cdot 4,200 / 6 \cdot 4,320 / 6 \cdot 4,4 / 10,16 / 10,32 / 10,64 / 10,125 / 10$. $200 / 10,2 \cdot 5 / 16,10 / 16,20 / 16,40 / 16,80 / 16,125 / 16,1 \cdot 6 / 25,6 \cdot 4 / 25,125 / 25,25 / 25,50 / 25$
$80 / 25,1 / 40,4 / 40,8 / 40,18 / 40,32 / 40,50 / 40,0.64 / 64,2 \cdot 5 / 64,5 / 64,10 / 64,20 / 64,32 / 64$. $80 / 25,1 / 40,4 / 40,8 / 40,16 / 40,32 / 40,50 / 40,0.64 / 64,2 \cdot 5 / 64,5 / 64,10 / 64,20 / 64,32 / 64$

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ELECTROLYTIC CAPACITORS Miniature P.C. mounting

## VEROBOAR

|  | $\times 3$ 3 | $\begin{aligned} & 01 \\ & 22 p \end{aligned}$ |
| :---: | :---: | :---: |
| 2 | $\times 5$ | 24p |
|  | $\times 37$ | 24p |
|  | $\times 5$ | 27p |
|  | $\times 2$ | 75p |
|  | $\times 34$ | 100p |
|  | $\times 5$ (plain) | - |
|  | $\times 37$ (plain) | - |
|  | $\times 2$ (plain) | - |
|  | $\times 5$ (plain) | - |
|  | $\times 3 \frac{1}{\text { (plain) }}$ |  |
| Pin | insertion too | 15 |
| Spot | face cutter | 42p |
| Pkt. | . 50 pins | 20 |

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Standard sereened $18 \mathrm{p} \quad 2.5 \mathrm{~mm}$ insulated Standard insulated $12 \mathrm{p} \quad 3.5 \mathrm{~mm}$ insulated Stereo screened 35p 3.5 mm screened Standard soened Stereo socke
D.I,N. PLUGS AND SOCKETS
D.I,N. PLUGS AND SOCKETS
2 pin, 3 pin, 5 pin $180^{\circ}, 5$ din $240^{\circ}, 6$ pin

2 pin, 3 pin, 5 pin $180^{\circ}$,
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| :--- | :--- | :--- |
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| $2500 \mu \mathrm{~F}$ | 40 V | 74 p |
| $2500 \mu \mathrm{~F}$ | 50 V | 58 p |
| $2500 \mu \mathrm{~F}$ | 64 V | 80 p |
| $2800 \mu \mathrm{~F}$ | 100 V | E 3.00 |

$3200 \mu \mathrm{~F}$
$3200 \mu \mathrm{~F}$
$4500 \mu \mathrm{~F}$ $4500 \mu \mathrm{~F}$
$1500 \mu \mathrm{~F}$ $\begin{array}{ll}500 \mu \mathrm{~F} & 25 \\ 5000 \mu \mathrm{~F} & 50\end{array}$ 16 V
16 V
25 V

50p
50p $50 p$
61.68
$\times 1.10$
$2500 \mu \mathrm{~F} 64 \mathrm{~V}$
$2800 \mu \mathrm{~F} 100 \mathrm{~V}$
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