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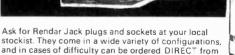
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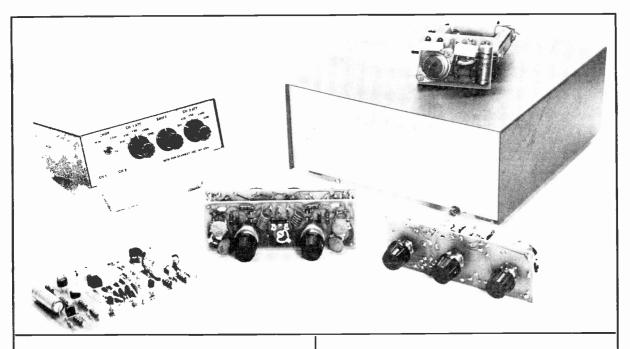
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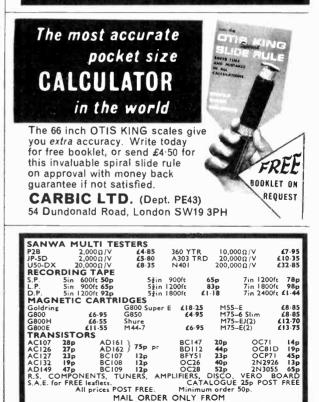
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venices, it covers the full medium and long wave bands. Four push-buttons for medium wave, one for long wave. It is permeability tuned and sturdily constructed. Output is a full 2.5 watts into an 8 ohm speaker. But the Tourist PB will operate into any loud-speaker from 8 to 15 ohms. Power consumption is less than 1 amp. The Tourist PB can be mounted in any standard size dash panel and it has an illuminated tuning scale for easy reading at night. Chassis size is: 7in wide, 2in, high and 4 18 in. deep (excluding front panel, etc.).

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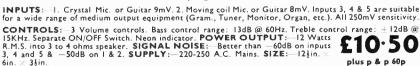




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PRIMARI Ref. Amps. No. 12V 24 11105 0:2 21310 0:5 71 2 1 18 4 2 70 6 3 108 8 4 72 10 5 17 16 8 115 20 10 187 30 15 226 60 30	5 12 7 1 0 8 1 0 7 2 4 8 3 12 10 5 4 10 5 3 7 7 8 12	6 × 5.7 × 4 3 × 5.1 × 5 0 × 6.4 × 5 3 × 7.0 × 7 2 × 7.6 × 8	4 0-12V a 1 0-12V a 7 0-12V a 0 0-12V a 2 0-12V a 2 0-12V a 2 0-12V a 1 0-12V a 1 0-12V a 1 0-12V a 1 0-12V a	$\begin{array}{c} t \ 0.25A \times 2 \\ t \ 0.5A \times 2 \\ t \ 1A \times 2 \\ t \ 3A \times 2 \\ t \ 3A \times 2 \\ t \ 5A \times 2 \\ t \ 5A \times 2 \\ t \ 10A \times 2 \\ t \ 10A \times 2 \\ t \ 15A \times 2 \\ t \ 30A \times 2 \\ \end{array}$	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$
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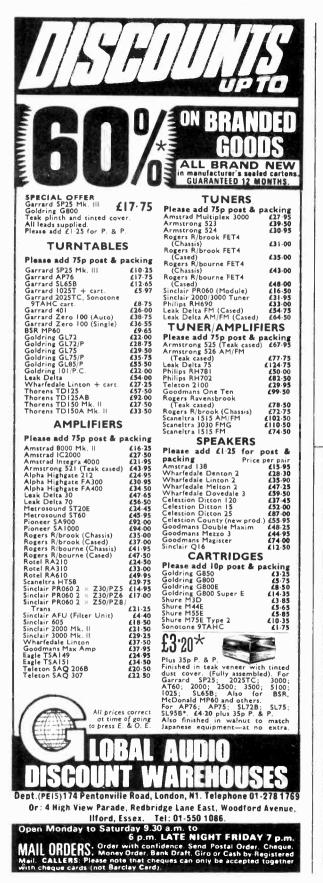
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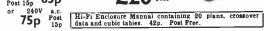
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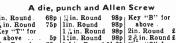


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AC107	0.20	AD162	0.83	BC148	0.10	BD137	0.45	BF188	0.40	OC19	0.35	2G371 -	0.16	2N2219	0.20	ubioor (			
VC113 VC115	02.0	AD161		BC149	0.12	BD138	0.50	BF194	0.12	OC20	0.63	2G371B		2N2219 2N2220	0.20	2N3054 2N3055	0.46	2N4059	0 10
AC117K	0.23	AD162 (		BC150	0.18	BD139	0.55	BF195	0.12	OC22	0.38	26373	0.17	2N2221	0.20	2N3391	0.14	2N4060 2N4061	0-12 0-12
AC122	0-12	ADT140	0.55	BC151 BC150	0.20	BD140	0.60	BF196	0.14	OC23	0.42	2G374	0.17	2N2222	0.20	2N 3391A		2N4061	0.12
AC125	0.17	AF114	0.24	BC152 BC153	0.17 0.28	BD155 BD175	0.80	BF197	0.14	OC24	0.56	2G377	0.30	2N2368	0.17	2N3392	0.14	2N4284	0.17
AC126	0.17	AF115	0.24	BC154	0 28	BD175 BD176	0.60	BF200 BF222	0.45	OC25	0.38	2G378	0.18	2N2369	0.14	2N3393	0.14	2N4285	0.17
AC127	0.17	AF116	0.24	BC157	0.18	BD177	0.65	BF257	0-95	OC26 OC28	0.25	2G381	0.16	2N2369.4		2N3394	0.14	2N4286	0.17
AC128	0.17	AF117	0.24	BC158	0.12	BD178	0.65	BF258	0.60	0C28 0C29	0·50 0·50	2G382 2G401	0.16	2N2411	0.24	2N 3395	0.17	2N4287	0 17
AC132	0.14	AF118	0 85	BC159	0.12	BD179	0.70	BF259	0.85	OC35	0.42	2G401 2G414	0.30	2N2412 2N2646	0.24	2N3402 2N3403	0.21	2N4288	0.17
AC134 AC137	0.14	AF124	0.30	BC160	0.45	BD180	0.70	BF262	0.55	OC36	0.50	26417	0.25	2N2711	0 21	2N3403 2N3404	0 · 21 0 · 28	2N4289 2N4290	0.17 0.17
AC141	0.14	AF125 AF126	0.25	BC161	0.20	BD185	0.65	BF263	0 55	OC41	0.20	2N388	0.85	2N2712	0 21	2N3405	0.42	2N 4290 2N 4291	0.17
4C141K	0.17	AF126 AF127	0.28	BC167 BC168	0·12 0·12	BD186	0.65	BF270	0.35	OC42	0.24	2N388A	0.55	2N2714	0.21	2N3414	0.15	2N4292	0.17
VC142	0 14	AF139	0.30	BC168	0 12	BD187 BD188	0.70	BF271 BF272	0.80	OC44	0.15	2N404	0 20	2N2904	0.17	2N3415	0.15	2N4293	0.17
AC142K	0.17	AF178	0.50	BC170	0 12	BD189	0.75	BF272 BF273	0-80 0-35	OC45 OC70	0.12	2N404A	0 - 28	2N2904A		2N3416	0.28	2N5172	0.15
AC151	0.15	AF179	0.50	BC171	0.14	BD190	0.75	BF274	0.35	0C71	0 10	2 N 5 2 4 2 N 5 2 7	0 42	2N2905	0 21	2N3417	0.28	2N5457	0 32
AC154	0.20	AF180	0.50	BC172	0.14	BD195	0.85	BFW10	0.60	0C72	0.14	2N598	0.42	2N2905 A 2N2906	0.15	2N3525 2N3646	0.75	2N5458	0.32
AC155 AC156	0 20	AF181	0.45	BC173	0.14	<b>BD196</b>	0.85	BFX29	0.27	OC74	0.14	2N599	0.45	2N2906A		2N3646 2N3702	0·09 0·10	2N5459 28301	0.40
AC156	0.20	AF186 AF239	0.45	BC174	0.14	BD197	0.80	BFX84	0.22	OC75	0.15	2N696	0 12	2N2907	0.20	2N3703	0.10	28301 28302A	0.50
AC165	0.20	AF 239 AL102	0.87	BC175 BC177	0.22	BD198	0.90	BFX 85	0 30	OC76	0.15	2N697	0.13	2N2907A		2N3704	0.11	28302A	0.42
AC166	0.20	AL102	0.65	BC177	0·19 0·19	BD199 BD200	0.95	BFX 86 BFX 87	0 22	0077	0 25	2N698	0.24	$2N292_{0}$	0.14	2N3705	0.10	28303	0-55
AC167	0.20	ASY26	0.25	BC179	0.19	BD205	0.80	BFX88	0.24	0C81 0C81D	0.15	2N699	0.35	2N2924	0.14	2N3706	0.09	28304	0.70
AC168	0.24	ASY27	0.80	BC189	0.24	BD206	0.80	BFY50	0.22	0C81D	0·15 0·15	2N706 2N706A	0.08	2N2925	0.14	2N3707	0.11	28305	0.84
AC169	0.14	ASY28	0.25	BC181	0.24	BD207	0.95	BFY51	0 20	0C82D	0 15	2N708	0 09 0 12	2N2926 (	0.12	2N3708	0.07	28306	0.84
AC176	0.20	A8¥29	0.25	BC182	0.10	BD208	0.95	BFY52	0.20	0C83	0.20	2N711	0 30	2N2926 (	(V)	2N3709 2N3710	0.09	28307 28321	0.84
AC177 AC178	0-24	ASY50	0.25	BC182L	0.10	BDY20	1 00	BFY53	0.17	OC84	0 20	2N717	0.35	2112020 (	0·11	2N3711	0.09	28321	0.56
AC179	0.28	ASY51 ASY52	0.25	BC183 EC183L	0.10	BF115	0.24	BPX 25	0.85	OC139	0.20	2N718	0.24	2N2926 (		2N3819	0.28	28322A	0.42
AC180	0.17	ASY54	0.25	BC183L BC184	0.10	BF117 BF118	0 · 45 0 · 70	BSX19 BSX20	0.15	OC140	0 20	2N718A	0.20		0.10	2N3820	0.50	28323	0.56
AC180 K	0.20	ASY55	0 25	BC184L	0 12	BF119	0.70	BSY25	0·15 0·15	OC169 OC170	0 25	2N726 2N727	0.28	2N2926 (		2N3821	0.35	28324	0.70
AC181	0.17	ASY56	0.25	BC186	0.28	BF121	0.45	BSY26	0.15	0C170	0.25	2N727 2N743	0.28	2N2926 (	0.10	2N3823	0.28	28325	0.70
AC181K	0.20	ASY57	0.25	BC187	0.28	BF123	0.50	BSY27	0.15	OC200	0.25	2N744	0.20	2242926 (	0.10	2N3903 2N3904	0.28	28326	0.70
AC187 AC187 K	0 · 22 0 · 20	ASY58 ASZ21	0.25	BC207	0.11	BF125	0.45	BSY28	0.15	OC201	0.28	2N914	0.14	2N3010	0.70	2N3904 2N3905	0.30	28327 28701	0.70
AC188	0.20	ASZ21 BC107	0-40	BC208 BC209	0.11	BF127	0.50	BSY29	0.15	OC202	0.28	2N918	0 30	2N3011	0.14	2N3906	0.27	40361	0.40
AC188K	0 20	BC108	0.08	BC209 BC212L	0·12 0·11	BF152 BF153	0.55	BSY38 BSY39	0.18	OC203	0 25	2N929	0.21	2N3053	0.17	2N4058	0.12	40362	0.45
ACY17	0.25	BC109	0.10	BC212L BC213L	0.11	BF153	0.45	BSY +0	0·18 0-28	OC204 OC205	0.25	2N930	0.21						
ACY18	0.20	BC113	0.10	BC214L	0.14	BF155	0.70	BSY41	0.28	OC205 OC309	0.35	2N1131 2N1132	0 20						
ACY19	0 20	BC114	0 15	BC225	0.25	BF156	0.48	BSY95	0.12	P346A	0.20	2N1302	0.14		DIOL	DES AND I	RECTIF	IERS	
ACY20 ACY21	0.20	BC115	0.15	BC226	0 85	BF157	0.55	BBY 95 A	0.12	P397	0.42	2N1303	0.14	AA119	0.08	BY133	0.21	OA10	0.35
ACY22	0 16	BC116 BC117	0.15	BCY30	0.24	BF158	0.55	Bu105	2.00	OCP71	0 43	2N1304	0.17	AA120	0.08	BY164	0.50	OA47	0.07
ACY27	0 18	BC118	0.10	BCY31 BCY32	0.26	BF159 BF160	0-60 0-40	C111E C400	0.50	ORP12	0.43	2N1305	0.17	AA129	0.08	BYX 38/3	0	0A70	0.07
ACY28	0.19	BC119	0.80	BCY33	0 22	BF162	0-40	C400	0.25	ORP60 ORP61	0 40 0 40	2N1306 2N1307	0 · 21 0 · 21	AAY30	0.09		0.42	OA79	0.07
ACY29	0.85	BC120	0.80	BCY34	0.25	BF163	0.40	C424	0.20	ST140	0.12	2N1307 2N1308	0.21	AAZ13 BA100	0.10	BYZIO	0.35	0.481	0.07
ACY30	0.28	BC125	0.12	BCY70	0.14	BF164	0-40	C425	0.50	ST141	0.17	2N1309	0.23	BA116	0 10 0 21	BYZ11 BYZ12	0.30	0A85	0.09
ACY31 ACY34	0.28	BC126	0.18	BCY71	0.18	BF165	0.40	C426	0.85	T1843	0.30	2N1613	0.20	BA126	0.22	BYZ13	0·30 0·25	0A90 0A91	0.06
ACY35	0·21 0·21	BC132 BC134	0 12 0 18	BCY72 BCZ10	0 14	BF167	0.22	C428	0.20	UT46	0.27	2N1711	0 20	BA148	0.14	BYZ16	0 40	OA95	0.07
10736	0.28	BC134 BC135	0.12	BCZ10 BCZ11	0 20	BF173 BF176	0 22	C441 C442	0.30	2G301	0.09	2N1889	0.82	BA154	0 12	BY217	0.35	0.4200	0.06
\CY40	0.17	BC136	0.15	BCZ12	0.25	BF170	0.35	C442 C444	0·30 0·35	2G302	0.19	2N1890	0.45	BA155	0.14	BYZ18	0.35	OA202	0.07
ACY41	0.18	BC137	0.15	BD121	0.60	BF178	0.30	C450	0.22	2G303 2G304	0·19 0·24	2N1893 2N2147	0-37 0-72	BA156 BY100	0.13	BYZ19	0 · 28	SD10	0.02
ACY44	0.32	BC139	0 40	BD123	0.65	BF179	0.30	MAT100	0.19	2G304	0.40	2N2147 2N2148	0.72	BY100 BY101	0·15 0·12	C(162 (Ea) 0 10		SD19	0.02
A D 1 30	0.38	BC140	0.30	BD124	0.90	BF180	0.80	MAT101	0.20	2G308	0.35	2N2160	0.60	B1 101 B1 105	0.12	(Eg) 0.49	0.05	1N34 1N34A	0·07 0·07
A D140 A D142	0.48	BC141 BC142	0.80	BD131	0.50	BF181	0.80		0.19	2G309	0.35	2N2192	0.35	BY114	0.12	CG651	0 00	1N914	0.07
AD142	0.38	BC142 BC143	0.80	BD132 BD133	0.60	BF182 BF183	0 40 0 40	MAT121 MPF102	0.20	2G339	0.50	2N2193	0.85	BY126	0.14	(Eq) 0A70		1N916	0.06
	0 50	BC145	0.45	BD135	0.40	BF184	0.40	MPF102 MPF104	0 42 0 87	2G339A 2G344	0.18	2N2194	0.35	BY127	0 15	OA79	0.08	IN414B	0.06
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BCY70	14p	13p	NKT243	510	44p	2N1711	150	14p		
BCY71	200	190	NKT401	700	56p	2N2904	290	28p		
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BFY50	190	180	OA79	60	50	2N3054	490	47p		
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2N1637 80p 2N1638 27p	2N 3855A	800 40	0412 50	BC213 Dp BC214	L 15p	BSY27 BSY28	159 OC35 179 OC36	50p 60p	CA3064 1809 949 TAA293 979 FCH101 859 MFC4000P TAA300 1759	6F6G 35p E180F 100p PL81 50p 6F13 45p EABC80 35p PL82 45p
2N1639 27p 2N1701 162p				P BCY10		BSY29 BSY32	170 OC41 250 OC42	227	FCH111 105p 55p TAA310 125p	6F14 70p EAF42 85p PL83 45p
2N1711 24p	2N3858	859 40	0528 71	BD BCY31	809	BSY36	25p OC44	25p 15p	FCH121 1059 SN7400 209 TAA320 729 FCH131 509 SN7401 209 TAA350 1759	6F15 65p EB91 20p PL84 40p 6F18 50p EBC41 55p PL500 75p
2N1889 887 2N1893 877	P 2N 3858A		0600 <b>5</b> ' 0603 <b>5</b> '	7p BCY32 0p BCY33		BSY37 BSY38	259 OC45 209 OC46	12p 15p	FCH141 105p 8N7402 20p TAA435 147p FCH151 105p 8N7403 20p TAA521 132p	6F23 85p[EBC81 30p PL504 80p
2N2147 78p	2N 3859A	82p A	C107 8	Op BCY34	80p	BSY 39	28p OC70	159	FCH171 1059 BN7404 209 TAA522 360p	6H6 17p EBF80 40p PY32 55p 6J4 50p EBF83 40p PY33 68p
2N2160 57p 2N2193 40p				DP BCY38 DP BCY39	40p 60p	BSY43 BSY51	50p OC71 82p OC72	12p 12p	PCH181 105p 8N7405 20p TAA530 495p FCH191 105p 8N7406 80p TAA811 445p	6J5 259 EBF89 839 PY80 409 6J5GT 809 EBL21 609 PY81 809
2N2193A 42 2N2194 27	2N3877	40p A	C128 2	Dp BCY40		B8 Y52 B8 Y53	82p OC73	20s I	FCH201 130p SN7408 20p TAB101 97p	6J6 20p EC86 60p PY82 35p
2N2194A 80p	2N3900	87p A	C152 2	8p BCY41 2p BCY42	: 15p	B8Y54	40p 0C75	309 22p	FCH211 1809 8N7409 209 TAD100 1509 FCH221 1809 8N7410 209 TAD110 1509	6J7 45p EC88 60p PY83 38p 6K8G 40p ECC40 65p PY88 40p
2N2217 25 2N2218 20	2N3900A			BP BCY43 Op BCY54			90p OC76 45p OC77	28p 30p	FCH231 150p 8N7411 28p 8L403D 150p	6L6GT 45p ECC84 30p PY800 40p
2N2219 20	2N 3903	20p A	C187 2	5p BCY58	289	B8 Y90	57p OC78	209	FCJ101 160p 8N7412 48p 8L702C 147p FCJ111 150p 8N7413 30p UA702A 280p	6LD20 50p ECC85 40p PY801 50p 6Q7 40p ECC88 40p U25 80p
2N2220 25 2N2221 25				5p BCY59 7p BCY60		B8Y95A C424	12p OC81 15p OC81D	20p 20p	FCJ121 275p 8N7416 84p UA702C 77p FCJ131 275p 8N7417 84p UA703C 187p	68A7 409 ECF80 359 U26 809 68G7 409 ECF82 359 U50 409
2N2222 201 2N2222A 251	P 2N3906	25p A	CY18 2	BCY70	15p	C450 GET102	15p OC82	259	FCJ141 585n SN7420 90n UA709C 45n	68J7 40p ECF86 65p U52 85p
2N2297 80p	2N4059	10p A	CY20 2	Op BCY72	15p	GET113	20p OC83	15p 85p	FCJ201 1009 8N7423 519 UA710C 1259 FCJ211 2759 8N7427 489 UA716 1879	68K7 40p ECH21 57p U191 75p 68L7 85p ECH35 100p U281 40p
2N2368 15p 2N2369 15p				0p BCY78 0p BCY79		GET114 GET118	20p OC84 20p OC139	26p 25p	FCK101 430p SN7428 80p UA723C 100p	68N7 85p ECH42 75p U282 40p
2N2369A 15p 2N2410 42p	2N4062	12p A	CY28 1	P BCZ10	27p	GET120	25p OC140	82p	FCL101 2809 8N7430 209 UA730C 1609 FCY101 1029 8N7432 489 UA741C 809	68Q7 40p ECH81 30p U301 40p 6U4 65p ECH83 45p U801 180p
2N2483 <b>27</b> p	2N4248	15p A	CY40 2	79   BCZ11 99   BD112	50p	GET880	30p OC171	25p 80p	BRIDGE   50PIV 4A 40p	6V6G 255 ECL80 455 UABC80 409 6V6GT 329 ECL82 355 UAF42 555
2N2484 82p 2N2539 22p	P 2N4249	15p A	CY41 1	59 BD116 59 BD121	112p	GET887	20p OC200 22p OC201	40p	RECTIFIERS 100PIV 4A 50p	6X4 359 ECL83 709 UBC41 509
2N2540 22p	2N4254	420 A	D140 4	7p BD123	80p	GET890	227 OC202	60p 75p	PLASTIC 200PIV 4A 55p 400PIV 4A 65p	6X5G 30p ECL86 40p UBC81 40p 6X5GT 40p EF37A 120p UBF80. 40p
2N2613 35p 2N2614 30p				7p BD124 Rp BD131	60p 75p	GET896 GET897	229 OC203 229 OC204	409	ENCAPBULATED 600PIV 4A 70p	10C2 50p EF39 50p UBF89 35p
2N2646 40p	2N4285	17p A	.D161 34	5p BD132	809	GET898	88p OC205	75p	600PIV 1A         50p         50PIV 6A         45p           50PIV 2A         45p         100PIV 6A         55p	10F1 75p EF40 50p UCC84 49p 10P13 60p EF41 65p UCC85 40p
N2711 25p N2712 25p	2N4287	17p A	F109 4	5p BDY10 5p BDY20	) 105p	MAT101	85p OC207	95p 75p	100PIV 2A 50p 200PIV 6A 65p 200PIV 2A 55p 400PIV 6A 75p	10P14 1109 EF42 709 UCF80 559 12AT6 809 EF60 859 UCH21 609
N2713 27p N2714 80p	2N4288	16p A	F114 24	BDY6	1 125p	MAT120	85p OCP71 85p ORP12	42p	400PIV 2A 80p 600PIV 6A 85p	12AT7 80p EF85 85p UCH42 70p
N2904 20p	2N4290	120 A	F116 2	50 BF115	25p	MJ400 3	107p ORP60	50y 40y	SILICON RECTIFIERS	12AU7 80p EF86 80p UCH81 40p 12AX7 80p EF89 28p UCL82 85p
N2904A 25p N2905 25p	2N4291	15p A 15p A	F117 20	0p BF117 0p BF152		MJ420 MJ421	80p ORP61 80p P346A	429	MINIATURE WIRE ENDED PLASTIC SERIES IN PL CL	12AV6 40p EF91 30p UCL83 60p
N2905A 20p	2N4294	17p A	F121 8	0p BF154	209	MJ430	1089 ST140	169	1 AMP 1-5 AMP 8 AMP	12BA6 409 EF92 359 UF41 609 12BE6 409 EF183 359 UF80 359
N2906A 85p		47p A 15p A	F125 1	BF158 BF159	255	MJ440 MJ480	95p 8T141 97p T1834	209 629	4002 100 PIV 7p 9p 20p	12BH7 45p EF184 85p UF85 40p 19AQ5 85p EH90 40p UF89 40p
N2907 28p N2923 15p	2N4965	18p A 52p A	F126 1	BF163 BF167	859	MJ481	125p TI843	40p	4003 200 PIV 89 109 289 4004 400 PIV 89 109 259	20D1 50p EL34 50p UL41 65p
N 2924 15p	2N 5028	579 A	F139 🙎	8p BF170	239	MJ491 3	100p TI844 187p TI845	10p 27p	4005 600 PIV 10p 12p 26p	20F2 659 EL41 609 UL84 409 20L1 1109 EL42 659 UY41 489
N2925 15p N2926G 10p		47p A 42p A	F178 41 F179 41	8p BF173 5p BF177	199	MJE340 MJE370	509 TI846 809 TI847	11p 11p	4006 800PIV 12p 15p 27p 4007 1000PIV 15p 16p 30p	20P1 000 PT.81 55m UY85 400
N29260 10p	2N5172	18p A	F180 👪	0p BF178	859	MJE371	80p TI848	189	50+ less 15% 100+ less 20%	20P4 110p EL84 25p VR105/3085p
N2926Y 10p N3011 20p	2N5175	52p A 52p A	F186 ا	0p BF179 0p BF180	350	MJE521	609 T1849 709 T1850	12p 12p	BILICOW RECTIFIERS STUD MOUNTING	20P5 1209 EL85 439 Add 129 in 8 25L6 509 EL91 859 for postage
N3014 82p N3053 18p	2N5176	40p A	F239 8	0p BF181 7p BF182	82p 80p	MPF102	437 T1851 359 T1852	10p 11p	6A 10A 17-5A 35A	DIODES & RECTIFIERS
N 3054 46p	2N5245	45p A	F280 4	7p BF184	201	MPF104	87p T1853	227	100PIV — 45p 50p \$1.82 200PIV 85p 50p 55p \$1.42	IN34A 10p; BA154 12p; GJ7M 37p
N3055 609 N3133 809	2N5246 2N5249	43 A 67 p A		8p BF185 5p BF194	20p 15p	MPF105 MP83638	879 X B112 829 X C141	139 359	400PIV 80p 55p 62p \$1.77	1N914 7p BAX13 12p OA5 17p 1N916 10p BAX16 7p OA6 13p
N3134 80p	2N5265	325p A	8Y27 \$4	0p BF195	169	NKT124	429 ZTX107	169	600PIV 889 609 729 48-18 800PIV 869 769 879 49-47	AA119 7p BAY31 7p OA9 10p
N3136 <b>25</b> p	2N 5305 2N 5306		8Y29 🙎	89   BF196 79   BF197	159	NKT126	879 ZTX108 879 ZTX109	159	1000PIV 409 859 \$1.05 \$2.77 50+ less 15% 100+ less 20%	AA129 100 BAY38 150 OA10 250 AA213 100 BY100 150 OA47 80
2N3390 <b>25</b> p	2N5307	879 A	SY50 SI	5p BF198	15p	NKT128	87p ZTX 300	180		AAZ15 109 BY103 229 0A70 79
2N3391A 80p		87p A 63p A	8Y54 2	BF200 BF224	14p	NKT187	879 ZTX 301 889 ZTX 302	90m	ZENER DIODES 400MW   1.5WATT   10 WATT	BA102 80p BY124 15p OA79 7p
2N3392 17p 2N3393 15p	2N5310		8¥67 44 8¥86 31	BF225 BF237	199	NKT210	30p ZTX 303	\$0p	3-3-33V 2-4-100V 3-9-100V	BA110 359 BY126 129 OA81 89 BA111 379 BY127 159 OA85 79
2N 3394 15p	2N5355	279 A	8Z21 51	BF238	- 28p	NKT212	309 ZTX 500	159	10p each 25p each 40p each 25+ less 15% 100+ less 20%	BA112 70p BY164 68p OA90 7p
2N3402 28p 2N3403 28p	2N 5356 2N 5365	829 A 479 B	UY10 150 C107 10	BFW6	289 1 479	NKT213 NKT214	200 ZTX 502	205	TRAFFISTOR DISCOUNTS :- 12 + 10%;	BA115 7p BY210 35p OA91 7p BA141 33p BY211 30p OA95 7p
2N3404 82p	2N5366	827 B	C108 10	Dp BFW8	7 259	NKT215	220 ZTX 503	179	25 + 15%; 100 + 20% any one type. Postage on all Semi-Conductors 7p extra.	BA142 88p BYZ12 80p 0A200 7p
	2N6367	579 B	C109 10 C113 10	BFW8 BFW8	8 <b>55</b> 7	NKT216 NKT217	859 ZTX 504 409 ZTX 531	409	S.A.E. FOR FULL LIFTS	BA144 139 BYZ13 359 0A202 109 BA145 309 BYZ16 409 0A210 179
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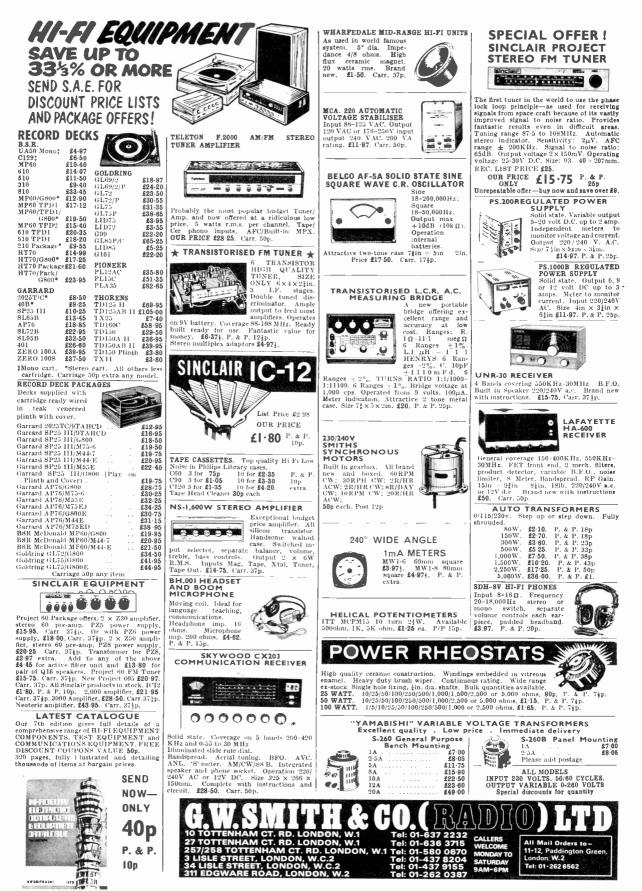
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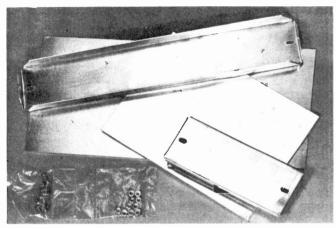


Practical Electronics

November 1972

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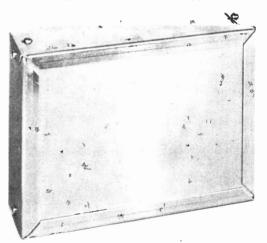
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## PRACTICAL VOL. 8 No. 11 **TRONIC** 1972 November

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## **OUR BUSINESS**

HIS magazine has its own established editorial conventions relating to constructional projects. These have remained unaltered since we started business and we have no plans to depart from these well proven conventions in the future. Perhaps these conventions should be clarified for the benefit of our readers.

Firstly, let us deal with the "exceptional". From time to time a constructional project appears with the prefix "PRACTICAL ELECTRONICS" incorporated in its title. This distinction is not awarded haphazardly. Our name is attached to selected projects only when the following conditions are fulfilled: the project is unique, in that it offers the home constructor something not previously available in a detailed comprehensive form, and it represents a significant or important new application of electronic circuitry. The original concept for the project may have come from the staff of this magazine; or may be the result of consultation with an outside designer. In either case, the actual development and production of the first model is put in the hands of a specialist contributor.

Apart from such commissioned designs, there are examples where a completed project has been offered to us in the normal way (without prior collaboration in the initial planning stage) and upon examination the project proves to "qualify" in the terms stated above. In such an event it may be decided to "adopt" this design to bring it into a special, close relationship with this magazine. This is subject to the agreement of the contributor naturally-but none so far has raised any objections to an adoption proposal!

We are certain that every one of the selected designs published under the P.E. insignia has been a credit to the designer concerned; furthermore we believe these designs stand out as notable landmarks in the eventful story of home constructed electronics, an activity which has progressed and expanded so dramatically over the recent years.

This special singling out of certain projects in no way diminishes the merit of the other projects published in these pages. All designs are rigorously vetted before acceptance and many of these do indeed indicate a very high degree of technical inventiveness and skilful workmanship on the part of the individuals responsible for their conception and subsequent execution.

Constructional projects are obviously the life blood of this magazine. Only projects designed and built expressly for the home constructor (the designer may himself be either a professional or an amateur in the electronics sense) are considered for inclusion under this continued on page 922

## THIS MONTH

## **CONSTRUCTIONAL PROJECTS**

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## SPECIAL SUPPLEMENT

ELECTRONIC MUSICAL INSTRUMENTS

Our December issue will be published on Friday, November 10

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THE audio frequency signal generator to be described here produces sine waves or pulses over the frequency range 19Hz to 200kHz in four bands. The amplitude of the sine waves is sensibly constant over the entire range, and the waveform good. The pulses are square in shape, with 1:1 mark/ space ratio over the lower three bands, that is from 19Hz to 20kHz. Above this frequency the waveform and level progressively deteriorates.

Simple

The maximum output level, both for sine and square waves is 3V peak-to-peak, just over one volt r.m.s. A constantly variable output control enables this to be reduced to zero, and is calibrated down to 100mV.

There is also a second output socket permanently attenuated by 20dB, thus permitting accurate setting of the output level to 10mV.

The complete unit operates from a self-contained 9V dry battery, and is housed in a case measuring  $7in \times 5in \times 2\frac{1}{2}in$ . Costing less than £4 to build from new components, this test set will be of value to the radio experimenter and hi fi enthusiast.

### WIEN BRIDGE

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PROMINEY (Ex)

The circuit is illustrated in Fig. 1. Transistors TR1 and TR2 are connected as an inverted Wien bridge oscillator. To sustain oscillation positive feedback is arranged between the two emitters via network LP1 and R5.

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By D. W. Easterling

2.0

WINNIP POLIS

At the same time, considerable negative feedback from TR2 emitter to TR1 base prevents the circuit from oscillating at any frequency except that selected by the Wien bridge network. This consists of VR1a, R1, VR2, R2, and VR1b; a pair of capacitors selected by the range switch S1a ganged to S1b completes the network. The variable resistors VR1a and VR1b are in fact one double-gang control which is used to provide continuous tuning over each range.

The series resistors R1 and R2 are chosen to provide a 10:1 frequency swing without permitting the minimum resistance of the network to reach such a low value that it seriously loads the emitter circuit of TR2.

The potentiometer VR2 enables both arms of the network to be balanced, particularly at the high

frequency (low resistance) end of each range, otherwise there is a tendency for the output level to dip beyond the auto-control circuit. VR2 can be omitted and the values of R1 and R2 set by trial. For example, in the prototype the values of R1 and R2 were 4.7 and 5.1 kilohms respectively.

The capacitors C1 to C8 determine the frequency ranges as follows:

Capacitors	Nominal Value	Frequency Range
C1, C2	120pF	19kHz to 200kHz
C3, C4	1,500pF	1.9kHz to 20kHz
C5. C6	0∙015µF	190Hz to 2kHz
C7, C8	0·15µF	19Hz to 200Hz

It will be noticed that C1 and C2 are made less than the expected 150pF, due to the effect of stray capacitances. To ensure accurate range multiplication each of these capacitors may consist of a 100pF silver mica fixed capacitor in parallel with a 30pF air spaced trimmer which is adjusted during calibration. In the prototype it was found that single fixed 120pF capacitors of close tolerance provided sufficient accuracy.

## POSITIVE FEEDBACK

The amount of positive feedback in the oscillator is critical. It must be sufficient to sustain oscillation without being so large that it drives the transistors off the linear part of their characteristic and produces unwanted harmonic distortion.

Unfortunately the changing characteristics of the load, power supply, and frequency selective network, tend to cause large variations in signal level which can only be overcome by using some form of automatic level control. Changes in level due to range switching can be countered by switching in different series resistors (VR3 to VR6), but this is insufficient when, as in this case, the frequency swing on each range is in the order of 10:1, consequently a more sophisticated arrangement is employed.

The usual method of solving this problem is to use sensitive thermistors which are both fragile, and costly. The circuit in Fig. 1 achieves the same result by using a 6 volt 40 mÅ lamp. The lamp in series with a  $150\Omega$  resistor is connected across the oscillator output.

As the output voltage increases, so does the current through the lamp, heating the filament and causing the lamp resistance to rise. This has the effect of reducing the voltage appearing at the junction of the lamp with the resistor, and thus the positive feedback. With careful design, the circuit becomes self balancing, and the output level stable.

It will be noticed that the lamp normally runs just below the level where it produces light, although under certain conditions a glow can just be seen. The nominal current through the lamp is in the order of 20mA.

The output of the oscillator is taken direct from the emitter of TR2 as a sine wave, or via TR3 and TR4 as a pulse or square wave. TR3 is a grounded emitter amplifier with a high value collector load resistor. Squaring occurs because the input signal is sufficient to bottom the base and collector characteristics on alternate half cycles.

Resistor R8 isolates the input circuit of TR3 from the output of TR2, thus preserving the sine wave output, and with R9, regulates the input to TR3 to ensure a square wave mark space ratio of 1:1. Adjustment of these resistors provides a convenient

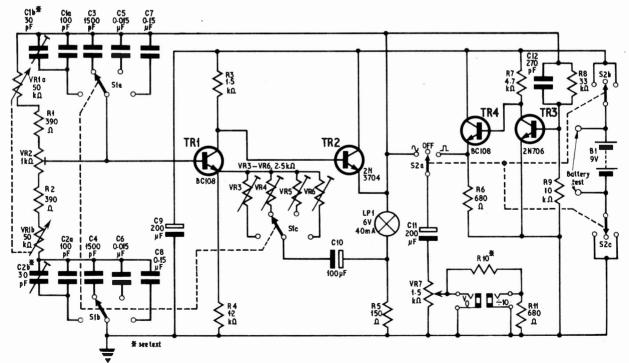


Fig. 1. The circuit diagram of the complete a.f. signal generator

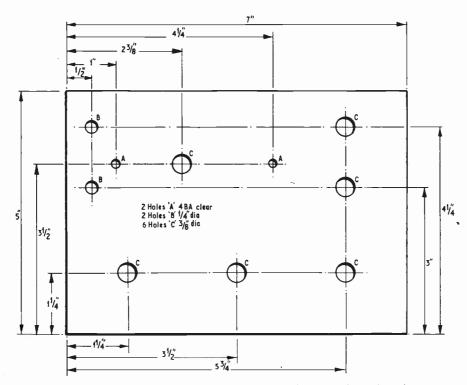
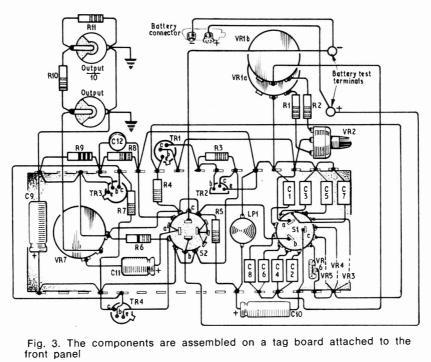


Fig. 2. The front panel drilling details to take the controls and sockets. The component board is bolted to this panel before the panel is covered with adhesive plastics sheet



## COMPONENTS . . .

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Capacitors         C1, C2         100pF silver mica in parallel with a 30pF air spaced trimmer or single 120pF silver mica (see text)         C3, C4         1,500pF 400V         C5, C6         15,000pF 400V         C7, C8         0.15µF 250V         C9         200µF elect. 16V         C10         100µF 400V         C11         200µF elect. 16V         C10         100µF 400V         C11         200µF elect. 16V         C12         270pF ceramic disc         C11         C11
<b>Transistors</b> TR1 BC108 TR2 2N3704 TR3 2N706 TR4 BC108
Switches S1 3-pole, 4-way switch S2 3-pole, 3-way switch
Battery B1 9V dry battery
Lamp LP1 6V 40mA lamp
Miscellaneous Aluminium case 7in × 5in × 2¼in Jack sockets (2 off) Wander sockets (2 off) Battery connector Pointer knobs Lamp holder, m.e.s. Component assembly board

method of tailoring the mark space ratio to individual requirements.

Capacitor C12 provides a degree of high frequency compensation. Transistor TR4 matches the high load resistance of TR3 to the comparatively low impedance output circuit.

### **OUTPUT TERMINALS**

Selection of sine wave or pulse is by switch S2a, which is ganged to S2b and S2c in series with the battery to put the unit to "off" in the centre switch position.

The standing d.c. potentials at the emitters of TR2 and TR4 are blocked from the output by capacitor C11, which must be of sufficiently large capacitance to pass the lowest frequency without significant attenuation.

The output control VR7 can be reliably calibrated down to 100mV, but for many applications, such

as testing pre-amplifiers, this is too high; hence the alternative output attenuated by network R10 and R11 which enables an output level of 10mV to be selected. Separate output sockets are preferred to a switched attenuator, as they facilitate simultaneous connection to two circuits, e.g. a test load and a monitor.

Terminals connected across the battery and brought out to the front panel facilitate battery checking, and may perhaps be useful in providing a power source for some other item of equipment but should not be allowed to interfere with the oscillator stability. The normal voltage across these terminals will be from 7.8 to 9 volts depending on the state of the battery. The battery should be replaced when the voltage falls below 7.8V on load.

### CONSTRUCTION

The aluminium case was purchased ready made and consists of two "chassis" type box trays with butting corners. The shallow tray acts as a lid and becomes the front panel. All of the components are mounted on it including the battery, as shown in the photograph.

The frequency control VRI dominates the front panel with sufficient space around it to accommodate a large pointer knob and associated calibration. It is positioned off-centre in order to provide accommodation for the battery which is mounted alongside VRI, and retained by spring curtain wire stretched between two 4B.A. bolts.

Switches S1 and S2, and the output control VR7, are mounted in line across the lower part of the panel. Their fixing also retains the tagboard on which all the other components including the transistors and preset controls are mounted. This arrangement facilitates wiring, which is kept as short as possible to minimise stray capacitance effects.

The holder for lamp LP1 is also wired on to the tagboard between S1 and S2. The output jacks and battery test terminals are fitted at opposite ends of the front panel in line with VR1.

In order to provide a suitable background for the calibration marking and switch legends the front panel is covered in self adhesive plastics sheet. The same material is used on the underside to insulate the tagboard. The panel is covered after it has been drilled and before the components were finally mounted.

A single sheet covered the whole panel including the holes, and unwanted parts were removed with a half round file. The covering will take instant lettering transfers, typewriter ink, or ball pen markings, which are then made permanent by a coating of clear varnish. Fingernail varnish is ideal. Without this protection the marks are easily removed using a clean cloth.

All the required words and figures can be typed on to a single sheet of plastics which is then varnished. It is a simple matter to cut out the appropriate word or figure as a label and attach it in the required position on the panel during the test and calibration process.

### TESTING

After carefully checking the wiring, and correctly connecting the battery, switch the unit on to sine waves. For this first test a current meter was inserted in series with the battery to monitor the current, which should not exceed 25mA. With the output connected to an oscilloscope, the appropriate positive feedback preset potentiometers VR3 to VR6 were adjusted for each range so that the level at the output remained sensibly constant over the entire sweep of the frequency control without there being any apparent distortion. The aim is to produce a clean sine wave with minimum harmonic content.

Due to the thermal lag of the lamp LP1, the level tends to "bounce" if the frequency control is adjusted too rapidly. A tendancy for the change in level between the low and high frequency ends of the frequency control to be outside the control circuit was corrected by adjustment of VR2.

Having established that the sine wave generator was operating satisfactorily, the unit is switched to square waves, which should be good up to 20kHz, but thereafter deteriorate as the frequency is increased.

As an alternative to an oscilloscope, headphones or an amplifier and loudspeaker could be used for audible testing of the unit over the three lower ranges. The higher frequencies are of course outside audibility. The positive feedback preset potentiometers should be adjusted to the point where oscillation is sustained over the entire range. Remember that too much feedback will produce a distorted waveform. On sine waves the tone is clear and pure. On square waves the harmonics become noticeable, the tone harsh, and apparently louder.

## CALIBRATION

Frequency calibration is most easily done using a counter. The writer used an oscilloscope to compare the low frequency end of the output against the 50Hz mains supply (via a suitable low voltage isolating transformer), and the high frequency end against a separate 100kHz oscillator beating against the carrier frequency of the transmission (BBC Radio-2 Long Wave 200kHz 1.500 metres) for the high frequencies.

An alternative method would be to compare the note heard through headphones with that from a well tuned piano. Piano keyboard frequencies based on orchestral pitch (A = 440Hz) are given in various reference books and cover the range 27.5Hz (bottom A) to 4,186Hz (top C).

The output control was calibrated in volts peak to peak because this value is meaningful for square waves as well as sine waves. Again an oscilloscope was used. An uncalibrated oscilloscope could be used for this purpose with reasonable accuracy by setting it up against the a.c. voltage range of a multimeter fed by the unit switched to sine waves at about 50 to 100Hz. One volt r.m.s. on the meter is  $1 \times 2\sqrt{2} \simeq 2.8$  volts peak to peak.

Suppose that the oscilloscope is adjusted so that 2.8V peak to peak produces a deflection of 5.6cm, then the oscilloscope sensitivity is 5.6/2.8 = 2cm/volt, and so 500mV peak to peak would produce a 1cm deflection. Lower values can be measured by expanding the trace. For example, with 500mV peak to peak producing a deflection of 5cm, 100mV peak to peak would produce a 1cm deflection.

The values of the fixed attenuator network R10 and R11 are nominal preferred values and were selected on test to reduce the output to one-tenth using the methods of measurement just described.  $\bigstar$ 

## OUR BUSINESS continued from page 917

heading. We do *not* include commercial kits in this category. Not that we have anything against kits. Some of our best advertisers are in this business, and for the man in a hurry, there's a lot to be said in their favour. But we feel it would be a mis-use of our editorial pages to describe in detail how to assemble some such kit. Provision of this kind of information is the business of the kit suppliers.

We know our business: encouraging genuine enthusiasts to build and learn; and encouraging individual circuit designers and application innovators to produce novel and practical projects to sustain this absorbing hobby. The steady expansion of our wide ranging readership at home and overseas convinces us that this policy is the right one. The almost insatiable demand for original designs is a challenge, but one our contributors and ourselves eagerly and willingly accept.

As stated in the opening paragraph, no changes are contemplated. (Why interfere with a successful formula?) PRACTICAL ELECTRONICS will continue as an unrivalled source of original design and construction information—conceived and planned throughout with the private constructor first and foremost in mind.

F.E.B.

# POINTS ARISING

### NOUGHTS AND CROSSES GAME (Oct. 1972)

In Fig. 6 p. 834 (caption omitted), all contacts on wafer Sle should be blank except for a "1" against position 5 Instead of position 6 as shown.

The 10-pole 10-way switch can be made from "Makaswitch" components but if a ready made switch is purchased, all wafers being make-before-break, then one wafer can be converted to break-before-make as required by carefully trimming the wiper contact until it does not short two contacts as the position is changed.

### **COMBINATION LOCK** (Ingenuity Unlimited Oct. 1972)

The transistors can be almost any low power *pnp* switching types or general purpose types, so long as they are all the same.

## **TTL EQUIVALENCE GATE** (Ingenuity Unlimited Oct. 1972)

In the diagram the last gate before the output has two inputs. These inputs are connected to the outputs of the preceding pair of gates respectively; they should not be joined together.

In the truth table, both "difference" conditions should have a "1" output.

### TRANSELECTOR (July 1972)

The value of R6 is  $2\cdot 2 \ k\Omega$ .

## WHO'S WHO

**I.C. INTERCOM** in the October 1972 issue has been accredited to the wrong author. The author of this article is **J. LEWIS**, to whom we offer our sincerest apologies. We regret any inconvenience which may have occurred to Mr Lewis and to Mr R. A. Penfold, who is the real author of the *CR Bridge* in this issue.



#### **ORBITING OBSERVATORY**

The largest telescope that has ever been put into space is carried aboard the newest orbiting observatory OAO-C. It is to be named Copernicus in a tribute to Nicklaus Copernicus as the father of modern astronomy. This forms a fitting part of the celebration of the 500th anniversary of his birth.

The observatory weighs some 4,900lb and orbits at 460 miles altitude. The spacecraft is about ten feet long and the 32-inch reflecting telescope occupies the centre of the vehicle.

In an effort to save weight the mirror is made of fused quartz and has been reduced in thickness to save 250lb. The actual weight is 105lb as against a normal reflecting mirror which would amount to about 350/360lb.

To achieve the necessary rigidity and constant figure the thin main disc is supported by quartz ribs and since gravity is reduced there is a further safeguard by carrying the telescope in space. This type of construction has been put into operation before by amateurs in an effort to reduce costs using the normal glass.

The 32-inch primary mirror collects the light rays and directs them to a smaller mirror 3.9 inches in diameter and thence to a spectrometer. The received data is then telemetred to earth.

This telescope is designed for the observation of ultra-violet light which is absorbed in the atmosphere. By working at a level where the atmospheric effects are minimal very valuable observations can be made. This type of work has been continuing for some time but with much smaller telescopes.

In addition to the large telescope there are three smaller telescopes on board the main purpose of which Work on the detection of X-ray sources has been going on for some time and to date some 200 sources have been identified.

In the early days X-ray studies were made with the sounding rockets and later with the famous Explorer 42 perhaps better known as UHURU which was launched in December 1970. The observations confirmed that there were X-radiations from distant sources very much greater in amount than those from the sun. It is believed that these may come from the far distant parts of the universe.

The record of these orbiting observatories has been an extremely good one. Apart from the unfortunate accidents during the early stages of the series, with the first one failing three days after launch in April 1966, and the third one failing to attain orbit because of the malfunction of the shroud jettison mechanism, the programme for the four observatories in this project has had its rewards. Some 10,000 observations of celestial objects have been made including major observations of comets and in May 1972 the event of a supernova.

On board the OAO-C observatory will be a computer capable of storing about 1.024 commands from the ground. This gives comprehensive and fully automatic operation while within range of the ground station at Rosman, North Carolina.

Any information obtained by the *Copernicus* observatory will be freely available to researchers all over the world.

#### LAST OF THE APOLLO'S

The Apollo 17 spacecraft left its berth at vehicle assembly Cape Kennedy, Florida on August 28, 1972. This spacecraft, the last of the Apollo series will carry the sixth team to a landing on the moon.

team to a landing on the moon. This last mission for this period will commence on Wednesday, December 6 at 21.53 EST (02.53 GMT Thursday) for the 12-day round trip to the moon. Three days will be spent on the surface of the moon.

Making his first space flight will be 37-year-old Harrison H. Schmitt, Ph.D., who is the first geologist to take part in an *Apollo* mission. His companion on the lunar surface will be 40-year-old Eugene A. Cernan, a veteran of some 264 hours of space flight and one of which took him to within 8 miles of the moon. Cernan will also undertake specially selected tasks during the mission.

The two astronauts will set up the fifth automatic station on the moon. The instruments to be set up are varied and one of these is the tidal gravimeter or lunar surface gravimeter. This instrument has been under development since 1964 and will be used to test the theory of gravity waves. This is at the moment a most important point in understanding the nature of the forces which celestial objects exert on each other.

Another type of gravimeter named the traverse gravimeter will be carried on the lunar rover. With this instrument the astronauts will record the changes in the gravity that take place over the lunar terrain. Thus, the anomalies will enable comparisons to be made with earth gravity changes.

Explosive measurements with seismic instruments will show the profile which will help determine the physical structure of the moon. Also operating will be instruments to check the cosmic dust and meteoritic fall-out, to see what are the effects of erosion on the moon.

The spacecraft will be launched on a Saturn V rocket and will be the first to be launched after dusk.

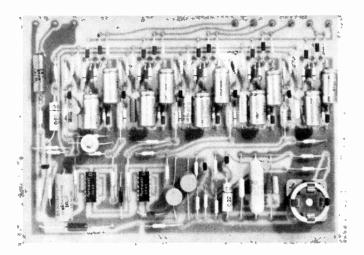
#### THE EXPLODED PLANET

Many people have explained the asteroid belt, the multitude of small bodies from less than egg size to chunks large enough for a space probe to land on, by the theory that there was a planet which was disrupted by the tremendous forces that exist in the solar system. This belt which lies between the orbits of Mars and Jupiter is being traversed by the Jupiter probe which may take photographs of some of them. In terms of proximity, of course, there is not likely to be many of these fragments near each other.

The problem of this debris is highlighted by the recent work of Michael Ovenden. Formerly of London and Edinburgh, Professor Ovendon is now at the University of British Columbia. He has been seeking a solution to certain celestial problems and his latest theory seems to be a satisfactory way of explaining the asteroid belt.

The various bodies which all interact with each other tend to settle to a minimum attraction position. Ovenden has checked this on such systems as Barnard's star which has planets. He has been able to get to an accuracy of 99 per cent with his theory.

To get the sums right for the earth and the rest of the solar system a planet 90 times heavier than the earth would be needed. It seems that the asteroid belt formed into a planet would be needed. It seems that the asteroid belt formed into a planet would satisfy this criterian. So he has called this planet Aztex. Though its real existence cannot yet be proved it is a good point from which to proceed.





#### Part 3

This part of the series is devoted solely to the main circuit boards of which there are 13 containing 90 per cent of the electronic components involved in the project. For convenient reference this board has been called the pitch board.

The function of the board, as described previously, is to generate a basic pitch and sub-harmonics of the same, followed by attack, decay, mixing, and tone circuits.

#### PITCH OSCILLATOR

The important requirements of the pitch oscillator is that it should have good long term frequency stability, good stability under varying ambient temperatures, and that it should present a suitable waveform which will reliably drive the frequency divider circuits. The circuit used is given in Fig. 3.1 which shows the various circuit elements that go to make up a complete pitch board.

The frequency of operation is determined by the tank circuit components C48 and L1. The inductance L1 is variable and is wound on a bobbin set in a ferrite core. After initial setting up C48 is fixed and comprises a number of capacitors in parallel dependent on the pitch concerned.

The oscillator transistor (TR16) is followed by a high gain buffer stage (TR17) which squares the waveform to give sufficiently fast edges at the collector to drive the first divider IC2. The cathode of diode D34 is returned to the divider supply rail in order to limit the positive excursion of the output waveform from TR17 ensuring that the maximum voltage presented to the divider input is approximately 5-3 volts.

#### PITCH DIVIDERS

In order to produce sufficient tones to cover the five octaves the fundamental frequency from the

buffer transistor is used for the top octave, whilst the other notes are produced at the outputs of four integrated circuit dividers which are packaged in

By A. J. Boothman B.Sc.

pairs. With the increase in availability of certain integrated circuits, the cost of these devices has dropped considerably, and to such an extent that in addition to circuit board area saving and ease of assembly, the incorporation of integrated circuits results in a significant cost reduction when compared with equivalent discrete circuity.

The divider circuits (IC2, IC3) are dual-in-line 14 lead packages each containing two flip-flops. The type of element used is described as a D-type edge triggered JK flip-flop. From its description it follows that the input waveform must have clean fast edges for reliable operation. For second, third and fourth stages this is provided by the previous flipflop, but for the first stage it is important that the output from TR17 should have fast edges. To ensure that this condition is met R119 is kept deliberately low.

#### **FLIP-FLOP ACTION**

The D-type flip-flop element (Fig. 3.1) operates in such a way that when a leading (positive going) edge appears at the clock input (CP) then the output at Q takes up the same state as that present at the D input immediately prior to the appearance of the clock pulse. Thus, if D was at a high voltage before the input signal rises then Q will take up the high voltage state and remain there after the input voltage falls again. If the input at D is then changed to a low voltage then Q will change to a low voltage as soon as the input waveform rises.

The state of the D input is controlled by externally connecting the  $\overline{Q}$  output to the D input. By definition Q is the complement of Q and so it can be seen that when the clock input is low D is always

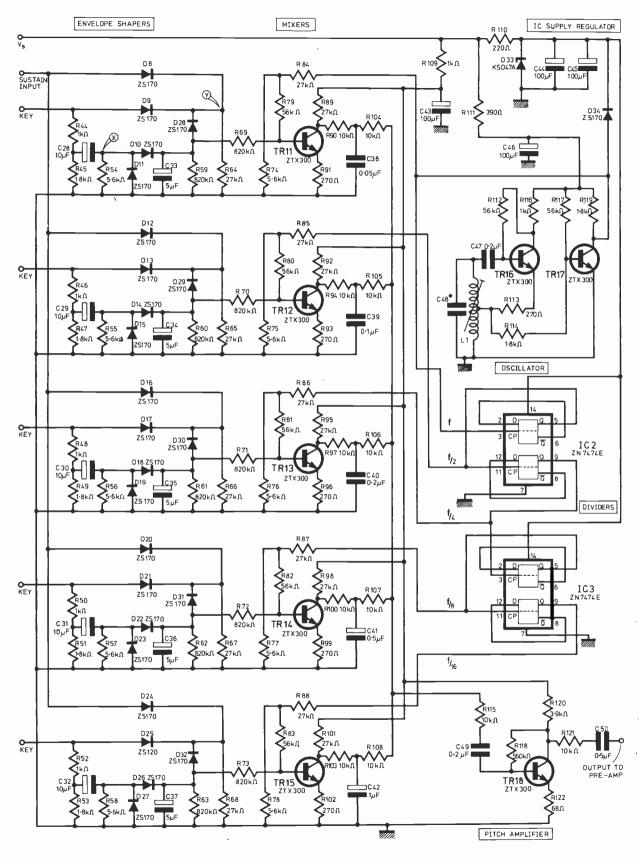


Fig. 3.1. Circuit diagram of a pitch p.c.b.

the opposite to Q, such that when the clock input rises Q will always change state, and remain in the new state until the clock has both fallen and risen again, at which point Q will again change state.

Since the clock input signal from the pitch generator is a square wave signal f, the Q output from the first flip-flop will be another square wave signal at half the frequency, f/2.

The other sub-harmonics are similarly generated by routing appropriate Q outputs to succeeding flipflops.

#### ATTACK AND DECAY ENVELOPE

The most important area in determining the character of this instrument is that concerned with the envelope, or attack and decay characteristic.

In order to obtain a percussive effect a fast rate of rise is required on the leading edge, followed by a rapid decay to a medium level of sound, and further followed by a relatively slow decay. On releasing the key the decay of sound should be fast but in the case of a piano this does take a finite time.

A string piano is activated by a hammer system where the level of attack is determined by the impact velocity of the hammer on the string, which is proportional to the rate of acceleration of the finger placed on the key. In the case of the electronic piano, in addition to a simulation of the hammer velocity action on each key, the maximum level of attack is determined by a five position touch switch which presets the key voltage supply.

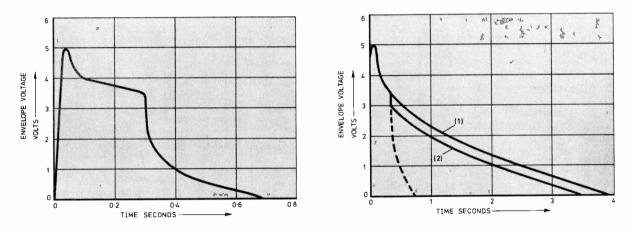


Fig. 3.2(a). Envelope circuit output characteristic in the non sustain mode with key released after 0.3 seconds. (b) Envelope circuit output characteristic with (1) key held depressed in the system or non sustain mode; (2) key released after 0.3 seconds in the sustain mode

The influencing factors involved are listed as follows:

- (a) The rate of rise of the leading edge of the envelope (attack).
- (b) The height of the leading edge.
- (c) The amount of rapid fall following the attack.
- (d) The length of decay when a key remains depressed.
- (e) The length of decay after the key is released (non-sustained).
- (f) The length of decay after the key is released with the sustain pedal on.

It is of course possible to display the output waveform of a string piano on an oscilloscope, and this could be followed by a point by point attempt at electronically simulating the envelope. In experimenting in this way, combined with straightforward comparative ear tests, the author noted certain features of the electronic equivalent which made it desireable to depart from an identical copy of the string piano envelope.

#### SLOW DECAY ENVELOPE

The curve shown in Fig. 3.2a indicates the envelope required under non-sustaining conditions, i.e. when a key is depressed and released fairly quickly without use of the sustain pedal.

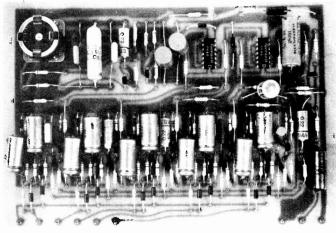
The level of initial attack on a string piano, when compared with the mean level of sound output, is very high, and a similar ratio is difficult to reproduce electronically if economical amplifier and loudspeaker capacities are to be used. This is the first point of compromise which has to be considered. Having recognised the need for a clear pip at the commencement of the envelope it is then necessary to consider the rate of decay in the medium level of sound region mentioned above. This is clearly coupled with the full decay period and again due to peak/mean power considerations it is somewhat slower than on a string piano.

Finally, when the key is released on a string piano a damper comes into operation immediately and quickly kills the vibration of the string. This action is reproduced realistically by an analogous electronic circuit.

#### LONG DECAY ENVELOPE

The curve shown in Fig. 3.2b indicates the other two possible conditions.

- 1. When the key is held depressed for a considerable time.
- 2. When the key is released quickly, but the sustain pedal is depressed.
- In both cases the first part of the envelope is



A completely assembled pitch p.c.b.

required to be exactly the same as in Fig. 3.2a. If the key remains depressed then the output decays slowly to zero in a time that is easily varied by the suitable choice of one resistor value in each envelope circuit. In the prototype this provides a decay of approximately 3.5 seconds, which is somewhat larger than for a string piano.

In the key released, sustain pedal on condition, it is more important to have a large amount of rapid fall following the attack in order to retain the percussive characteristic of the instrument when repeatedly playing the same note at a fast rate. As mentioned earlier a string piano has a very high peak capability and a fast initial decay. This allows fast repetition of one note whilst, even under sustain conditions, maintaining the discrete identity of each impact via its attack characteristic.

The attack characteristic of the first curve in Fig. 3.2a is insufficient to reveal differentiation between consecutive notes in the sustain condition, and a second characteristic is therefore introduced whereby a second fall in level occurs as the key is released, followed by a shorter decay time of approximately 3 seconds.

#### THE ENVELOPE CIRCUIT

Each of the d.c. key inputs in Fig. 3.1 are shaped by envelope circuits. As can be seen there are five duplicated circuits that precede the mixers to fulfil this function. To understand the action of these let us look at the topmost shaper of Fig. 3.1.

A note is initiated when a d.c. level of between 11 and 18 volts, depending on the touch setting, is applied to the input of the circuit via the keyboard. Capacitor C33 is charged via R44 and C28, with R44 providing some slight slowing of the rise time in order to protect against key noise, and also being a factor in determining the height of the attack pip.

Following the relatively quick charging of C33, C28 is charged which means that point "X" falls towards ground potential. The presence of D10 isolates the input circuitry from C33 and there is no further interaction between the two circuits. Removal of the key voltage results in the discharge of C28 via the diode D11 and resistor R45.

It is important that R54 should be sufficiently low to provide fairly rapid charging of C28, and

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that R45 should be sufficiently low to allow complete discharge of C28 before the next key voltage is applied. However, both resistors are involved in determining the voltage applied to C33 during attack (voltage sharing with R44), and all values are therefore both important and interrelated.

#### TOUCH SENSITIVITY

In order to give touch sensitivity R45 can be replaced by a grounded switch contact in the keyswitch action which becomes a double throw switch between a zero volt potential and the supply voltage on the second pole.

During the interval when the switch is changing over, C28 is charged by a bleed resistor between rail voltage and the centre contact of the changeover switch. This reduces the attack voltage excursion at the point when the changeover is complete, by an amount dependent on the time taken over the movement between the two poles. Further details on this will be given later.

#### SUSTAIN AND DAMPING

Now, ignoring the input circuit, we can consider the sustain and damper portion of the envelope. The keying operation had charged C33 to approximately 5 volts, and whilst the key remains depressed point "Y" is held at approximately 14 volts, thus isolated from the output line by diode D28.

In this condition C33 is slowly discharged by the combined effect of R59 and the load placed across the output. If at any point during this decay the key should be released then D28 and R64 would appear across the output, and since R64 is made very low compared to R59 this provides the damper action.

When the sustain pedal is in operation D8 is held at approximately 4.5 volts which causes a quick drop in the output voltage on release of the key, but then holds off the action of R64 during the remaining part of the sustain period.

#### TONE FORMING

Tone forming in the system is very simple and consists only of a low pass capacitive filter after each transistor. The complete circuit of the combined envelope and tone forming section is shown in Fig. 3.1.

As stated in the first article the prime purpose of the project is to provide a reliable portable keyboard instrument with a percussive characteristic similar to that of the piano. The tone of a piano is of course extremely variable from one instrument to another, and whilst it is possible to incorporate involved tone circuits to match one particular piano colouring, the author decided that the simple filter above produced a sound which is completely acceptable for both the home and band applications.

#### TONAL CHARACTERISTICS

It is difficult for the author to give an objective description of the sound produced by the instrument, but generally the piano type characteristics are most noticeable above middle C, and the attack becomes less apparent in the lower registers, where a fuller more organ like sound is evident.

When driving an external amplifier of course, using the appropriate output socket, the tonal

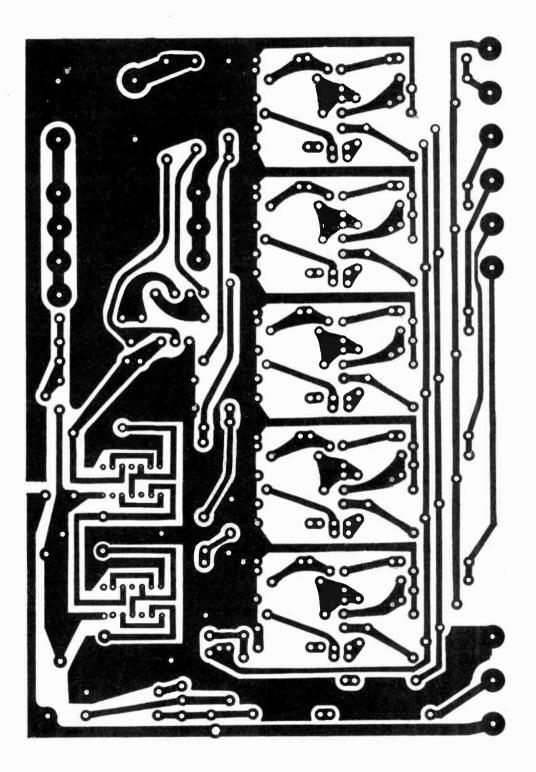


Fig. 3.3. Etching details for pitch p.c.b. This is full size

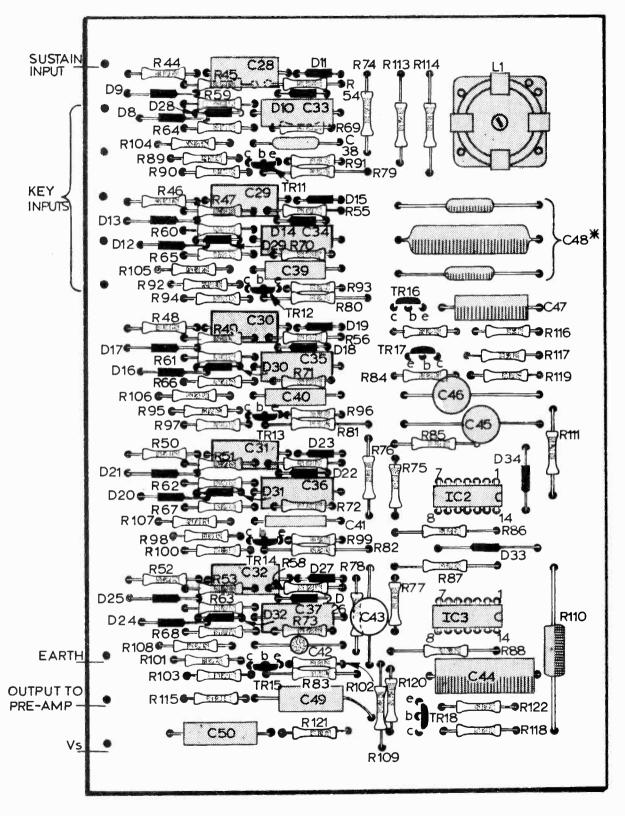


Fig. 3.4. Component layout for pitch p.c.b.

characteristics can be varied over a wide range by variation of the tone control settings on the external amplifiers.

The piano has been used in this manner on a number of occasions and exhibits the characteristics of piano, vibraphone and electric guitar depending on the chosen bass and treble positions, and the way in which it is played.

#### PITCH AMPLIFIER

Each printed circuit board contains five envelope circuits which are then fed into a pitch amplifier which in turn is followed by the pre-amplifier described last month.

The values of C49, R121 and C50 are important, as up to this point all coupling has been direct, and

it is necessary in the amplifier to filter out the large very low frequency excursions due to the keying action.

#### ETCHING DETAILS OF PITCH P.C.B.

The  $8in \times 5\frac{3}{8}in$  printed circuit board is etched according to the pattern of Fig. 3.3. Thirteen such boards are required. Fig. 3.4 shows the component layout for the boards.

#### COMPONENTS

All transistors specified are Ferranti type ZTX300. It is of course possible that other equivalent transistors could perform the same functions, but no work has been carried out to prove this. The diodes contained in the prototype were a mixed batch of

### COMPONENTS ...

#### PITCH BOARD

(12 are required with a full component complement as listed. The thirteenth requires components as shown asterisked)

Resistors		R118* 560kΩ
R44*	1kΩ	R119* 1·8kΩ
R45*	1-8kΩ	R120* 3·9k Ω
R46	1kΩ	R121* 10kΩ
R47	1⋅8kΩ	R122* 68Ω
R48	1kΩ	All resistors $\pm$ 5% $\frac{1}{4}$ W carbon except where other-
R49	1·8kΩ	wise stated
R50	1kΩ	
R51	1·8kΩ	Capacitors
R52	1kΩ	C28*-C32 10µF elect, 15V (5 off)
R53	1-8kΩ	C33*C37 5µF elect. 15V (5 off)
R54*-R58	5.6kΩ (5 off)	C38* 0.05µF
R59*-R63	820kΩ (5 off)	C39 0-1µF
R64*-R68	$27k\Omega$ (5 off)	C40 0.2µF
R69*-R73	820kΩ (5 off)	C41 0.5µF elect. 15∨
R74*-R78	5.6kΩ (5 off)	C42 $1\mu$ F elect. 15V
R79*-R83	56kΩ (5 off)	C43* 100µF elect. 15V
R84*-R88	27kΩ (5 off)	C44 $100\mu$ F elect. 15V
R89*	27kΩ	$C45$ 100 $\mu$ F elect. 15V
R90*	10kΩ	$C45^{*}$ 100µF elect. 15V
R91*	270Ω	$C40^{+} = 100 \mu F$ elect, 13V C47* $0.2 \mu F$
R92	27632 27kΩ	$C47$ $O2\mu^{r}$ C48* See text
R92	270Ω	C40 See text C49* $0.2\mu$ F
R93	10kΩ	$C49^{-}$ $0.2\mu$ F C50* $0.5\mu$ F
R94 R95	27kΩ	C50 0.5µr
		Transistors
R96	10kΩ	
R97	270Ω	TR11-TR18 ZTX300 (8 off)
R98	27kΩ	(4 required for p.c.b. 13)
R99	270 \	
R100	10kΩ	Diodes
R101	27kΩ	D8-D32 Z\$170 (25 off) (see text)
R102	270 Ω	(6 required for p.c.b. 13)
R103	10kΩ	D33—KS047A 4·7V 400mW Zener (or equiv-
	10kΩ (5 off)	alent)
R109*	1kΩ	
R110	220Ω 3W	Integrated Circuits
R111*	<b>39</b> 0 Ω	IC2–IC3 Z7474E (2 off)
R112*	56kΩ	
R113*	270 Ω	Inductor
R114*	1·8kΩ	L1—260 + 260 turns of 36 s.w.g. enamelled copper
R115*	10kΩ	wire. S.E.I. pot core assembly for this can be
R116*	1kΩ	obtained from Clef Products (Electronics
R117*	56k $\Omega$	Division), Yew Tree Lane, Poynton, Stockport, Cheshire



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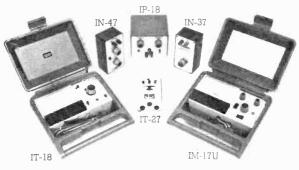
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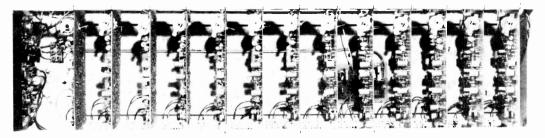
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The thirteen pitch boards mounted in a wooden sub-frame

rejects from the Ferranti ZS170 and ZS100 series production lines. Reject devices are not available from the manufacturer, but similar diodes are available from a number of suppliers.

The specification required is that all diodes should have a breakdown in excess of 20 volts, and be reasonable low leakage, which should be adequately covered by the use of silicon types.

The 4.7 volt Zener diodes were Ferranti type KSO47A, but again a number of equivalents should also be available.

#### PITCH BOARD LAYOUT

In arranging a component layout for the pitch board an attempt was made to match the physical to the circuit arrangement. This can be seen by comparing Figs. 3.1 and 3.4.

Thirteen boards are required in total. Of these 12 are required containing the full circuitry and they cover bottom F to top E.

It must be obvious, however, that in a five octave 61 key system a single note generator is required to produce top F. In general the only requirements for the thirteenth board are an oscillator, single envelope circuit, mixer and pre-amplifier. No divider circuitry is required.

The components required for this particular board are marked with an asterisk in the components list.

#### WINDING THE TUNING COIL

To make up the tuning inductors for the oscil-lators, simply pile wind 260 turns of 36 s.w.g. enamelled copper wire onto the bobbin. With even lavers this works out at about 33 turns/laver. When this is done bring the wire end out through one of the slots on the bobbin on the same face as entry. Doubling the wire back, continue to wind on an additional 260 turns. The point where the wire is doubled is, of course, the centre tap and this and the two ends should be cleaned prior to tag soldering.

To hold the coil in place on the bobbin, wrap round a piece of insulation tape. With the pot core clamped round the bobbin and lead outs soldered to the tag board the inductor can be assembled to the p.c.b.

Details on the tuning process and choice of capacitor will be provided later.

Note: All enquiries for Kimber-Allen keyboard should be directed to Alan Douglas, 4 Lees Barn Road, Radcliffe-on-Trent, Notts.



FOR the fourth consecutive year PRACTICAL ELEC-TRONICS is exhibiting at the International Audio Festival and Fair, held at Olympia, London from October 24 to 28, 1972. Equipment on display on Stand 19 includes the P.E. Electronic Piano (currently being described in the magazine) and a number of other exciting projects to be published in the coming months; these are:

#### Sound Synthesiser

The P.E. Music Synthesiser is a completely selfcontained sound effects unit whose composition possibilities extend as far as the user's imagination. The modular construction of the various electronic subassemblies means that the synthesiser can provide a limited facility for sound source production and treatments even in early stages of construction. This, of course, offsets the total production cost and enables the constructor to build, test and play with the separate modules according to his effects requirement.

A fascinating feature is that the synthesiser can be programmed to play by itself for as long as the listener wishes. (See Lecture-Demonstration below.)

**Rhythm Generator** 

This variable-tempo rhythm generator can be programmed to any required rhythm to accompany a piano, organ or other musical instruments.

The percussive effects available are bass drum, high and low bongos, blocks, snare drum, cymbals and long and short brushes. Besides the programmed facility a selector switch provides a choice of a number of popular rhythms.

#### **High Quality Tape Link**

Forthcoming articles will describe how to build new stereophonic electronics for an existing tape transport deck or for a complete recorder.

The Tape Link employs separate record and replay amplifiers, the latter being driven via an extremely low current bias f.e.t. stage from a low impedance head; consequently noise and distortion figures are kept to a very Replay equalisation is achieved with an low level. integrated circuit operational amplifier, which also provides frequency compensation for circuit stability and output short circuit protection. Power amplifiers are not included. Tape speeds catered for are  $1\frac{7}{8}$ ,  $3\frac{3}{4}$ ,  $7\frac{1}{2}$  inches per second.

#### Lecture-Demonstration

Visitors to the Audio Fair should note the P.E. Sound Synthesiser is being demonstrated during a special lecture entitled "Sound Synthesis for the Amateur", presented in the exhibition lecture theatre on Tuesday, October 24 and again on Saturday, October 28, at 2 p.m.

We welcome all readers in London in October to see our exclusive designs in the flesh. Only at Olympia, October 24 to 28, 1972.

**S** INCE few component manufacturers use really indelible identification markings on their components, it is inevitable that sooner or later the home constructor will require a piece of equipment, which will enable him to measure the value of a component from which these markings have been erased. Such a unit is also extremely useful when trying to locate a faulty component, or when trying to identify one with an unfamiliar colour coding. The bridge described in this article will fulfil all these functions.

#### CIRCUIT DESCRIPTION

An extremely simple circuit with only three transistors is used. Ranges of the bridge cover capacitance from 100pF to  $10\mu$ F and resistance up to 100 kilohms, in five ranges of capacitance, and five ranges of resistance.

These ranges are as follows:

Range	Capacitance	Resistance
1	100pf-1000pf	0-10Ω
2	$1000 \text{pf} - 0.01 \mu \text{F}$	0-10012
3	$0.01\mu F - 0.1\mu F$	$0-1k\Omega$
4	$0.1\mu F - 1\mu F$	0–10k()
5	$1\mu$ F $-10\mu$ F	0–100kΩ

#### THE WHEATSTONE BRIDGE

The circuit is based on the Wheatstone Bridge, and a short explanation of the operation of this will be given.

The circuit of a basic Wheatstone Bridge is given in Fig. 1. It will be seen that the bridge consists of four resistors and a meter. There will be no indication on the meter when the circuit is balanced, that

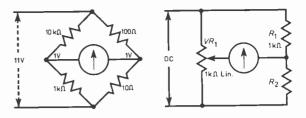


Fig. 1. The basic Wheatstone Bridge circuit (left)

Fig. 2. A modification of the Wheatsone Bridge circuit normally used for resistance measurement (right)

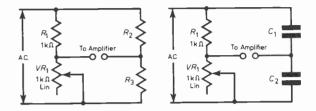
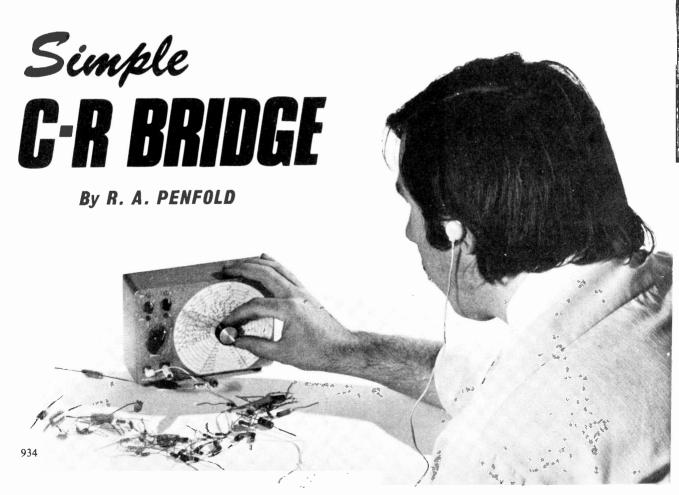


Fig. 3. An improvement of the previous circuit which produces a linear scale (left)

Fig. 4. A modification of the basic bridge circuit used to measure capacitance. This system does not produce a linear scale (right)



is, when there is the same voltage at the junction of the 1 kilohm and the 10 kilohm resistors as there is at the junction of the 10 ohm and 100 ohm resistors.

The meter will read the difference between the two potentials at these points. As there is one volt across both points, there is no difference in potential, and there can be no deflection on the meter.

Should one of these resistors be altered in value, even very slightly, then there would be an indication on the meter. If  $R_1$  and  $R_2$  were replaced by a potentiometer, as in Fig. 2, then it would be possible to adjust this component so that the circuit was once again balanced.

The voltage on the slider of the potentiometer would be variable from zero to the supply voltage, and could therefore be adjusted for zero reading on the meter, whatever the ratio of  $R_1$  to  $R_2$ .

If the variable resistor is fitted with a suitable scale, then when this component is adjusted for a null on the meter, the value of the resistor under test may then be read off the scale.

This is of course only possible if  $R_1$  is a fixed, and known value, or a series of resistors of known value. Whatever the value given to  $R_1$ , the same value will appear at the centre of the scale of  $VR_1$ .

#### IMPROVED METHOD

Theoretically the circuit is capable of measuring any resistance from zero to infinity, but in practise, the scale would become so cramped at the ends, that it would be impossible to use.

it would be impossible to use. Therefore if  $R_1$  is made of several switchable resistors, then the resistor to be tested can be measured near the centre of the scale, where it is less cramped.

An improved arrangement is shown in Fig. 3. This will measure resistance from zero to the value of  $R_3$  with a linear scale. There is no cramping of the

scale, which covers a far smaller range than the arrangement of Fig. 2.

This arrangement is also different in that the d.c. source has been replaced by an a.c. source, and the meter has been replaced by an audio amplifier, which would be connected to an earpiece.

Using an a.c. source enables capacitance to be measured by changing the resistor,  $R_2$ , for a capacitor. Fig. 4 shows the arrangement for the measurement of capacitance.

Unfortunately, when measuring capacitance a linear scale is not obtained. This is due to reactance (resistance to an a.c. flow) decreasing as capacitance increases.

This means that the bridge will measure from the value of  $C_1$  to infinity, with a centre scale of double the value of  $C_1$ . The scale does become rather cramped at the high value end, but this is far less so than the arrangement of Fig. 2.

#### THE PRACTICAL CIRCUIT

A circuit diagram of the complete C-R Bridge is given in Fig. 5.

A multivibrator using two BSY95A transistors is used to provide the a.c. source. The audio amplifier uses a single type 2N2926 transistor, which feeds into a high impedance crystal or magnetic earpiece.

The range switch is a two-pole five-way wafer switch which selects one of five resistors from 10 ohms to 100 kilohms and one of five capacitors from 100pF to  $1\mu$ F. A second switch is used to select either resistance or capacitance.

Separate power supplies have to be used for the amplifier and the multivibrator, to prevent the amplifier from picking up the output from the multivibrator through the supply lines. This also makes the coupling between the bridge, and the oscillator and amplifier much simpler.

Two torch batteries are used to supply the power.

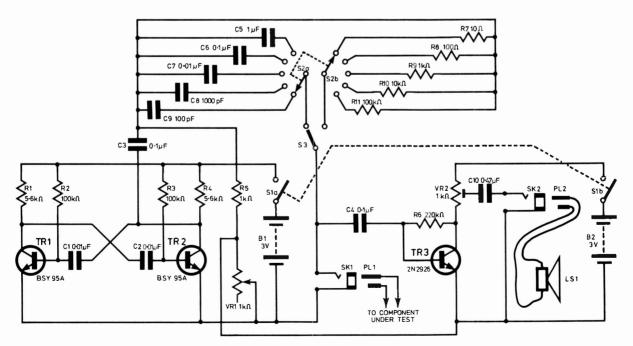


Fig. 5. Circuit diagram of the complete unit with the oscillator, bridge and amplifier

### COMPONENTS . . .

$ \begin{array}{c} \textbf{Resistors} \\ \textbf{R1} & 5{\cdot}6k\Omega \\ \textbf{R2} & 100k\Omega \\ \textbf{R3} & 100k\Omega \\ \textbf{R4} & 5{\cdot}6k\Omega \\ \textbf{R5} & 1k\Omega2\% \\ \textbf{R6} & 220k\Omega \\ \textbf{R7} & 10\Omega \\ \textbf{R8} & 100\Omega \\ \textbf{R9} & 1k\Omega \\ \textbf{R9} & 1k\Omega \\ \textbf{R10} & 10k\Omega \\ \textbf{R11} & 100k\Omega \\ \textbf{All} \pm 10\% \ \frac{1}{2} \textbf{W} \text{ unless otherwise stated} \\ \end{array} $
Potentiometers VR1 1kΩ linear wirewound VR2 1kΩ miniature skeleton preset
Capacitors         C1, C2 $0.1\mu$ F ceramic (2 off)         C3, C4 $0.1\mu$ F ceramic (2 off)         C5 $1\mu$ F         C6 $0.1\mu$ F         C7 $0.01\mu$ F         C8 $1000$ pF         C9 $100$ pF         C10 $0.47\mu$ F
<b>Transistors</b> TR1, TR2 BSY95A <i>or</i> similar (2 off) TR3 2N2926
Switches S1 2 pole on/off slide S2 2 pole 5-way rotary S3 1 pole changeover slide
Miscellaneous PL1, PL2 and SK1, SK2 Miniature jack plugs and sockets (2 off) LS1 High impedance earpiece B1, B2 3V torch batteries (2 off) 3≩in × 1≩in 0.15in matrix Veroboard Aluminium case, battery clips, crocodile clips

#### COMPONENTS

As the sensitivity of the earpiece may vary according to which particular type is used, a one kilohm preset potentiometer is incorporated, so that the output may be reduced, should it be found to be excessive. The circuit is unsuitable for use with a low impedance earphone.

Although adequate results may be obtained using a carbon potentiometer for VR1, better results will almost certainly be obtained using a wirewound type. Other components for the bridge should preferably be one or two per cent tolerance types, with a high stability, if a reasonable accuracy is to be attained.

Transistors other than the types specified have been found to work in the circuit, and it would operate using almost any transistors.

#### CONSTRUCTION

The prototype was built in an aluminium case, measuring  $6\frac{3}{4}$  in  $\times 4\frac{3}{4}$  in  $\times 2\frac{5}{5}$  in, but almost any case of about these dimensions could be used, although the unit could not easily be made much smaller than this. If a metal case is used, neither the oscillator or the amplifier should be earthed to it as this would encourage oscillator breakthrough.

A small piece of Veroboard (0.15in matrix) 21  $\times$  7 holes, was used as a basis on which to construct the main circuit. A wiring diagram for this is given in Fig. 6.

The best method for connecting the component under test into circuit was found to be a lead with a jack plug one end, and a couple of crocodile clips the other end.

The jack plug may be plugged into the bridge, while the crocodile clips connect to the leads of the component under test.

#### INTERWIRING

All wiring associated with VR1, switches S1 and S2, and the two crocodile clip leads, should be as short as possible otherwise breakthrough of oscillations may make it impossible to locate a proper null point on some ranges, and capacitance in the wiring could make capacitance Range 1 wildly inaccurate.

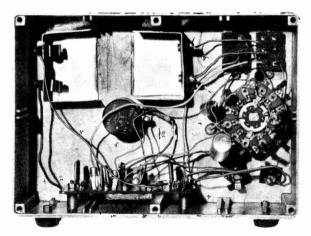
Both the earphone socket and the test lead socket have to be insulated from the front panel to prevent them from being electrically connected, and to prevent breakthrough. This may be achieved by drilling the mounting holes slightly too large, and then fitting bakelite washers either side of the panel.

Capacitor C10 is only required if the bridge is to be used with a magnetic earpiece. If a crystal type is used, C10 may be omitted, and the earpiece connected directly to the slider of VR2.

#### CALIBRATION ACCURACY

Although both the resistance and the capacitance scales can be calculated mathematically this is unnecessarily complicated, and is unlikely to be as accurate as the following method. This method is to calibrate the scale against a number of close tolerance components of suitable values.

An individual scale is required for each range. The resistance scales will read in the opposite direction to the capacitance scales. They should be marked in preferred values, from 10 to 100. At the end of each scale there should be added a multiplier. For example. On Range 4 resistance  $\times$  100 should be added at the end of the scale, so that if



Rear view of the unit showing the disposition of the batteries, component board and switches

If one goes back far enough, and includes isolated private experiments, it is difficult to know just when electrical music started it is certainly more than 100 years ago But practical instruments have only be existence for some 0 years, so it is a comparatively new art, now rapidly expanding with the semiconductor and its offshoots.

ELECTRONIC

MUSICA

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STRUMENTS

I F I I I I I L L L I I L Martin I

Electronics is linked to music in two ways : either the sound is produced by a conventional instrument which uses electrical means to process it in some way (e g amplification or change in tonal quality), but where the sound could still be heard there were no electronics; or, the "sound is generated by electronic circuits, originating as electrical oscillations or waveforms and therefore inaudible without further electronic processing.

Electronics is a science; music is an art, They do not speak the same language for one is based on provable parameters, whilst the other relies entirely on personal For this reason we cannot comment on the aesthetic appeal (or otherwise of electronic music. But to the technically inclined, there are other appeals; the future is pregnant with unlimited scope for generating new sounds, modifying existing ones, simplifying playing techniques and many other laces of

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Electronics and music have grown into a formidable combination and so we take this opportunity of bringing together some of the many examples of this association. But fully to appreciate these grout advances, which so many people take for granted, one should know something of the long and fascinating history of electronic musical instrument development; the more so as this is a hobbies journal, and some of the ideas which were not suitable for commercial development make interesting propositions for "one-off" projects.

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ELECTRONIC ORGANS

We begin with electronic organs, because these instruments represent the most ingenious combination of mechanics and electronics. can often be built at home, and are capable of endless develop-ment. All such instruments are played from keyboards and the controls follow internationally recognised methods to a great extent, but certainly not exclusively. The whole concept of organs has changed in the last 20 years. There is now no resemblance (in the popular mind) to the conventional organ with pipes. Yet, incredibly, the same conventional names for the sounds appear on the stops!

Harking back to the days when valve circuits tended to be unstable, mechanical systems with transducers seemed the hest answer; of these, two survive today, Hammond and Compton. Such organs are permanently in tune but require complex circuitry and switching devices. With the increasing cost of manufacture it seems probable that they will dis-appear in time. Already solid state circuitry is taking the place of switching in pipe mechanical organs, although slowly, but for other kinds of electronic organ, semiconductors are paramount.

It is easier to start from the top and work downwards, so we give an example of an imitative organ by Conn as representing the "real thing". Fig. 1 illustrates a church or concert organ having tonal and control properties exactly corresponding with pipes.

This instrument has independently tuned oscillators for each note and these can give at least three different waveforms for tonal synthesis. As the sound in a pipe organ comes from a widely dispersed source, so this class of electronic organ requires many channels of amplification and many loudspeakers—a minimum of perhaps 25.

Multiple oscillators are never exactly in tune, but perhaps more important is the fact that there are trifling differences in the harmonic content of each oscillator. This gives a close resemblance to organ pipes, which have the same properties. The sound is full and rich, excellent for a real organ, but quite useless for pop or jazz.

Nine out of ten organs of the popular domestic type which can be seen in any music shop use a quite different generating system. The 12 top notes are produced by tuned master oscillators, but all other pitches downwards are obtained by frequency division. Since the octave is an exact doubling or halving of any frequency, 2:1 bistables or flip-flop circuits serve admirably. Readers will recall that this was the method used in the P.E. organ. It lends itself to mass production especially now that i.c.s are becoming so cheap; and with diode or transistor keying, cheap single contacts per key suffice.

If properly designed, this kind of organ can be very satisfying more especially at the low output levels demanded in the home. Unfortunately it is commonly found in forms which have neither the proper playing facilities of imitative organs, or the "new" sounds associated with rhythmic organs. Excessive amplification makes these organs lose what character they possess, and underlines the old saying that you can't get a quart out of a pint pot.

However, the prompt attack and the easy application of rhythm units to this kind of instrument makes them eminently suitable for pop groups and this has led to a form of tonal synthesis which contrasts well with guitar or voice. To a great extent they must be treated

Fig. 1. The three manual Conn Custom organ has independently tuned oscillators for each note. These can provide at least three different waveforms for tonal synthesis of a church or concert pipe organ.

The Kentucky Challenger with string, flute and woodwind voicing available via the 4-octave solo and accompaniment manuals.

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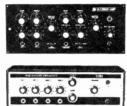
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DISCO-AMP. 100 wait rms. output for 8-16 ohms, 4 channel inputs. 2-mic, 2 decks. Separate volume control plus masters, Response 30Hz -30KHz, distortion less than 1%. Treble/Bass/ PFL/Mic over-ride etc. Panel size 164" × 7″

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all the necessary features for a high quality mo-bile unit. The pre-amp has separate tone con-trols for both mic, and decks, and each input has its own indi-vidual volume controls and inputs, plus the addition of a cross faile for deck to deck sound transfer. A built in P.F.L. system for cuelnet, together with mic-over-ride facility are standard on all units. Response 20-20,000 Hz. Mic. input 5mV. 50K Outnut L volt

are standard on all units. Response 20-20,000 Hz. Mic. input 5m v. 56K. Output i volt. McDonaid M.P.60 Turntables are used with high quality ceramic cartridges, and each deck has its own individual cut out switch fitted. This unit is suitable for Discos or Clubs having a power amplifier, or for use with the Disconster' 100 watt power amplifier as above. Size  $32^{\circ} \times 20^{\circ} \times 8^{\circ}$ , **298-50**.



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Designed to take three E/S Type spot or colour bulbs up to 150 watts each. The unit is of all metal construction and has one 3-pin mains input socket plus one 3-pin mainoutput socket for connecting more than one bank together if required. The unit can be left free-standing or wall mounted if needed. Black crackle finish gives the unit a very professional appcarance.

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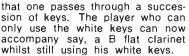
as new sound devices and we show one of the most modern and effective examples in Fig. 2. This is made by Jennings Electronic Instruments, and bristles with ingenious features.

In this organ, the emphasis is on facilities for rhythmic playing rather than on simulation of conventional organ tonecolours; a special training and a special mentality is required to conform to the needs of a purely rhythmic group, and this is an excellent example of an instrument specially designed to this end.

The foregoing organs have all had tuned oscillators for each primary function and once these are adjusted, it is not possible to alter them easily. But now, it is possible to design and make an organ having only one master oscillator from which all pitches are derived. One great advantage is clearly that transposing is easily carried out—i.e., if one presses just one key, the pitch can be changed by a single adjustment so

Fig. 2. The three manual, portable, Jennings J.71 with drawbar control of harmonics from 16' to 1' pitch and percussive attack. This, with the provision of dual waveforms, enables the organist to reproduce any mixture of sound required.

Other features include separate by-pass tabs for Theatre and Baroque organ. The top manual covers three octave providing 16'-8'- and 2' pitch with controls for tonal effects which include vibraharp, piano, string, percussion and sustain.



There are several ways of doing all this, mostly very complex, and all a tribute to microminiature circuits; for without these it would be quite impossible to build such an organ within any console or case. Modern i.c. component density can reach 2,000 elements in a chip only lin square and so the very complex circuitry required to steer and smooth the pulses of such a system can be made quite small. Two of the present methods for generating the frequency spectrum from a single source have been described in Practical Electronics for July 1972 but in the picture of the Philips generator the actual size was not obvious; in Fig. 3 you can see the relative size of this unit

An organ built from half a dozen of these units would have amazing powers of synthesis.

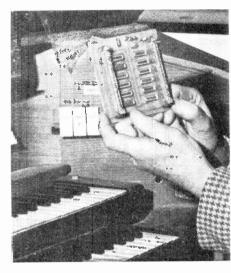


Fig. 3. Demonstrating the small size of the N.V. Philips' digital organ generator.

One of Hammond's new models, the Concorde, incorporates custom buil! LSI packages. This harmonic tone-bar organ has two 61 note keyboards and such features as polysyntoesis percussion, manual and pedal sustain, automatic accompaniment and automatic rhythm, lower manual to pedal couples and cassette recorder.





Talking about synthesis, we find small keyboard controlled devices mounted in the Wurlitzer organs on which many effects not obtainable with any ordinary organ cir-cuits can be produced. For instance, gliding tones, wah-wah, touch sensitive vibrato, extended pitch range and other effects are some of the attributes of the Orbit synthesisers on this organ. Certainly it is time that double touch was more used on organs, especially since it was an integral fitment of all cinema organs since the 1920s. So many things can be done with a second touch.

Now we have seen three kinds of tone generating systems for organs of a more or less conventional kind. But for the beginner. or one who does not wish to lay out much money on a single-purpose instrument, there are other ways of making musical sounds. Perhaps the simplest device is the

Fig. 4. The Dubreq Stylophone has a reedy voice, derived from a multivibrator circuit. To this vibrato can be added if desired. This instrument is only capable of producing one note at a time, each being selected with a metal stylus which constitutes part of the electronic circuit. In concert the sound produced can be very pleasant.

stylophone, shown in Fig. 4. This is a simple multivibrator, played by using a metallic stylus to make the connection for changing the pitch. It is a continuously re-tuned system, capable of only one note at a time, but having a vibrato circuit which relieves the monotony. Constructional articles for something of this kind have appeared in various journals and it is a nice project for the beginner.

A more sophisticated device is the Pianomate. This assumes ownership of a piano, although it could be energised by a separate keyboard. It can be seen in Fig. 5 lying over the keys of a piano. Again, it is a simple generator, but chords can be played with care. The system uses one generator for every three notes, but if played with the piano, which of course is fully polyphonic, it appears that the organesque sounds from the generator are also fully polyphonic.

A small mains unit which stands on the floor contains all the electronics and allows changes of tonecolour; a volume control pedal is also available. Incidentally, as the contact unit for the four octaves of keys only rests on the back of the piano keyboard, it can be removed or placed on another make of piano in a moment and cannot damage the woodwork.

We find many people interested in playing music who are not satisfied with synthetic tone colours: certainly many organs become very trying to listen to for long and rely heavily on vibrato to break up the lifeless sounds. Many attempts have been made to bring the actual sounds from organ pipes or orchestral instruments to the keyboard. culminating in the Mellotron. Originally this instrument was much more complex but the present version is easily portable and just as accurate tonally.

Fig. 5. The Dubreq Pianomate in position on a conventional pianoforte keyboard. This consists of two double octave unit whose contacts move with the piano key to augment the piano sounds with organesque tone colours. A tone selector switch provides three different colours: flute, church organ and jazz buzz. A two speed vibrato is also included. The Pianomate is completely tuneable via a single control for pianos which may be out of pitch by as much as a semitone



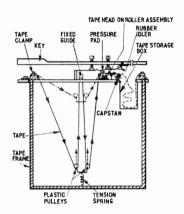
PRACTICAL ELECTRONICS SUPPLEMENT- NOVEMBER 1972





The system uses pre-recorded tapes but these are not in the form of loops as might be thought; the ingenious mechanism is too complex to describe here, but in essence when a key is pressed, a tape head is brought into contact with a record of the selected sound and continues to sound (so long as the key is held down) for up to 8 seconds. The tape then rewinds at once and is ready again. All the sounds provided are of course as authentic as the fidelity of the reproducer, all can be mixed, and many effects sounds can be brought in on other tapes; indeed, the Mellotron is popular as a pure effects machine, from which every conceivable noise can be obtained at will. Fig. 6 illustrates the action of this instrument.

Fig. 6 A pure effects machine, the Mellotron is a source for every conceivable noise. The system uses prerecorded tapes of other instrumental sounds or special effects which can be mixed at will by the performer making him, in effect, a one man band. The action of the Mellotron can be understood by referring to the profile drawing of the key and tape transport system;when the key is depressed the idler engages with the capstan pulling the tape into the storage box; the tape is kept in tension by the rising pulleys; with the key released the tape is made ready for replay as the tension spring returns to its original position pulling the stored tape with it.



The RIHA Largo, besides having the normal footages in the upper and lower manual, has fractional number stops to provide more colourful registration. Playing features include a two-speed Leslie tremulant (a spatial effect achieved by feeding the loudspeaker output into a rotatable drum), vibrato delay, which provides a much more natural vibrato sound and solo percussion which gives the choice of many effects such as banjo and Hawaiian guitar.

The sustain feature on the pedals provides a stringbass effect on the pedal 8' stops so giving rhythmic support to melody on the manuals. The sustain can also be used to play legato on the 13 note pedalboard. The Harmonics Solette organ with a specification designed to cover all aspects of organ music from the classics to pop. It has a full size 61 note keyboard and a total of 19 registers.



# Now that you've got AUG10 You've never heard it so good!



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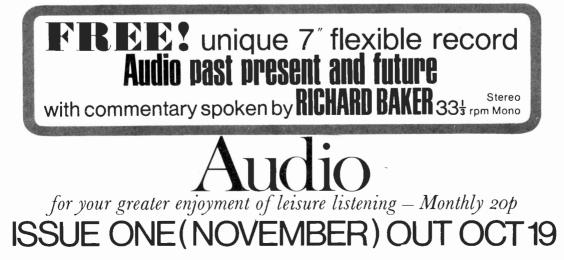
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## INSTRUMENT TRANSDUCERS

Electric guitar pick ups have been with us for a long time; more recently the use of transducers has spread to other instruments such as clarinets, saxophones, violins and accordions. The aim of these methods is twofold; to amplify the original sound, and to be able to alter the waveform of the sound by electronic means, e.g. to remove harmonics or to add extra octaves by frequency division; or to modify the steady tone in some other way. But firstly let us look at how the quitar operates.

The diameter of a string is so small compared with the wavelength of any frequency within the range of that string that the back wave from the vibration would at once neutralise the front wave and nothing would be heard; the string produces no sound, it drives the belly or resonating body of the guitar and this couples with sufficient air to make the resultant vibrations audible as sound. However, it is costly to make a good acoustic guitar and the resulting sound is not loud enough to compete with other sounds in a modern group; the acoustic guitar is therefore reserved for the classical player, since it is many times more sensitive to fingering than its electrical counterpart, and the tone is not coloured by amplifiers or loudspeakers.

For the pop scene, it is desirable to convert all sounds into electrical waveforms and hence have complete control over them. So all we need is the vibrating string with suitable pick-ups. This means a simpler instrument in theory, but by the time that eleccontrols and perhaps tronic multiple pick-ups are fitted, it may well be more complex. Construc-tional data for pick-ups has been published and it is very easy to amplify this sound; one can then add simple tone controls and so alter the effect. If, therefore, it is possible to use steel strings or strings wrapped with steel wire over the pick-up location, magnetic pick ups are easily applied to any stringed instrument.

Note that the string pick-up is to be preferred to any form of microphone attachment; because the latter is sensitive to air pressure waves as well, and can pick up breathing and scratching noises from fingering the string. But when one comes to other instruments, in which the same amount of energy is not required of the player (or in a different form, such as blowing a clarinet), then it is possible to use noise cancelling types of microphone to amplify the tone. Certainly this has advantages if only the waveform of the because generated sound can be changed

and one may get the effect of several different kinds of instrument from the one. An excellent example of this is the electric mouth organ or harmonica.

There are examples of amplified accordions and a number of purely electronic ones, using the circuitry of a miniature organ and connected to external amplifiers and loudspeakers. Then again, it is possible to obtain an electronic string bass unit, buttons producing all the effects from normal bowing to pizzicato or col legno (playing with the back of the bow). This is made by Hohner.

Many of these existing tone qualities can be completely transformed by, say, a percussion unit; this enables some sound which would normally be smooth and sonorous to be made aggressive and staccato. This kind of percussion unit can be wired in anvwhere between a pick-up and its amplifier. It is not to be confused with a rhythm unit, which is a quite independent source of percussive sounds acting as rhythm markers and non-musical in effect. Many such rhythm boxes are on the market now, nearly every large domestic organ has one as an integral part and many can be bought to add to whatever sound source would benefit.

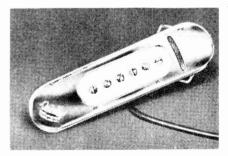
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Farfissa Transicord electric accordion with seven voices, sustain and vibrato tabs. A rhythm section can be used in conjunction with a bass-chord facility. It has a 41-note key and 120 bass buttons. A mains supply and pre-amplifier are contained in the separate module Hornby • Skewes accordion microphone is connected to the side of the accordion and picks up sound for feeding to an amplifier





Magnetic type contact pick-up with fingertip control. Suitable for nylon string guitar.

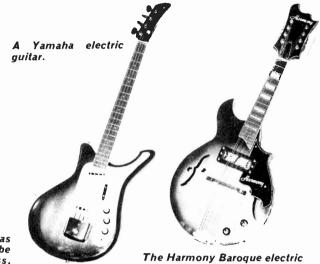


Six magnet pick-up for flat top guitars.



The Farfisa Super Piano with keyboard and decay as on the conventional piano. The effects which can be obtained on the first 24 notes are, for the bass: bass, string bass and bass guitar. For the chords: pianoforte, guitar and banjo. Also available is automatic bass and chord rhythms. Output available is 25 watts.





The Harmony Baroque electric mandoline with a body shape that departs from the classical for ease of playing.



The Jennings Winchester "Rifle" guitar with a body which is constructed entirely of metal providing a crisp, pure tone. Apart from the normal guitar sound this instrument has internal electronics to provide bass, treble, fuzz, presence and repeat.



Typical reverberation unit from the Laney Sound Supergroup Series.



A "Cry-baby" wah-wah pedal.



"Fuzz-face" distortion unit with separate volume and fuzz controls.



Soundimension: a device for providing echo and reverberation affects when used with an amplifying system.

## SPECIAL EFFECTS

The continual search for new effects has led to the electronic controls known as wah-wah, growl, glide, etc. All of these do something not normally expected from the instrument to which they are attached. Once associated solely with the guitar, they are now found on electronic organs and other complex devices. Readers will be familiar with the effects on the tone, and constructional details for most have been described in the press.

The wah-wah is a tuned circuit, the resonance of which can be altered at will; some band of frequencies is then accentuated and moves progressively to another band by the manipulation of a foot control. The growl is almost identical, but operates at lower frequencies. A glide circuit is found on some organs, usually it provides a limited frequency shift by altering the base biasing of a transistor oscillator through a voltage control; again the foot is used to move a variable resistor of some form.

Since some of these effects may be needed at short notice, it is possible to combine the controls on a single foot pedestal where the toe selects the function, whilst the heel operates a rotary volume or similar control. A commercial multiple control unit is shown in Fig. 7. There is clearly a limit to the number of controls which can be used by someone who is usually playing an instrument at the same time.

The vibrato produced by a violinist, which is an important part of his technique, may be imparted by a mechanism which alters the tension of the guitar strings by a lever; but it can also be done by an electronic circuit—commonly a form of adjustable speed multivibrator. This again can be inserted between the pick-up and the amplifier but naturally it is not so expressive or controllable as manipulation of the strings. However, playing finesse is not so important with groups.

Since power is such an essential ingredient with today's performers, we find large amplifiers which are now sophisticated in that they have mixing and vibrato circuits incorporated; however, we shall not discuss these or loudspeakers at this time. Suffice it to say that many groups prefer valves, since they stand such overloads and misuse.

There is room for improvement in the means for connecting units together, there still are casualties from time to time and it is difficult to understand why Ministry of Defence or Home Office approved connectors are not compulsory.

One final piece of apparatus which has effects uses is the ring modulator. This is a circuit artifice whereby two applied signals are combined to form sum and difference frequencies, one of which is extracted and used as a final signal. For this reason the device should be fed from sine waves, but since these do not exist outside the laboratory, some very peculiar sounds result—always discordant.

The ring modulator is often fed from a musical source, a singing voice via a microphone, a pick-up from a saxophone, etc. Many intriguing effects are possible with care, certainly all are novel and often incapable of analysis by the hearing system yet of too short a duration to give rise to irritation. It is in fact interesting to note that distortion is deliberately sought, so intense is the search for novelty. The clipping amplifier or fuzz box is a good example of such techniques. Of course, as the ambient noise level around us increases, discomfort arising from other noises has less effect and we come to accept it.

Fig. 7. The Jennings Scrambler is a complete remote foot control of effects associated with amplification. Besides the quadrant of foot switches the Scrambler has two rotary turntable controls for intensity volume and wah-wah.

CRAMBLES

A quartet of effects units that can be used with almost any kind of electronic musical instrument.









# SYNTHESISERS

The idea of compounding sounds from their bare ingredients is far from new; indeed, one of the most successful synthesisers was built 20 years ago; but, it occupies a whole room! So for a more general acceptance of these devices, we had once again to wait for the semiconductor.

Now a synthesiser is nothing more than a number of units which represent the basic parameters of sounds, so organised that they are easily controlled by the operator and of such a nature that the absolute values of these parameters can be set up again and again with accuracy; in other words, so far as is possible, the elements are calibrated.

One can compose music with a range of tonal qualities, but can-

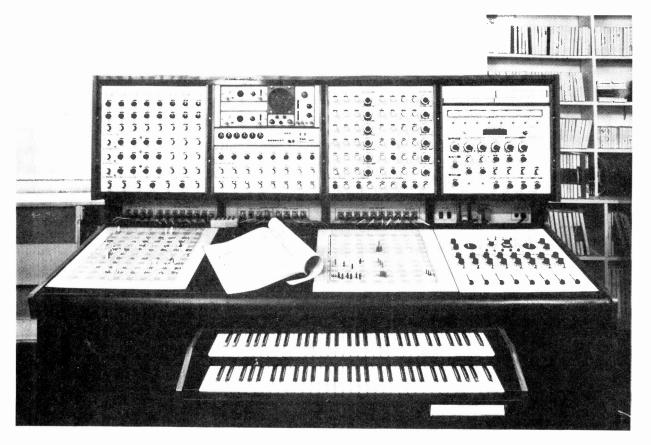
The Synthi 100 synthesiser made by EMS of London has almost unlimited facilities for sound synthesis and serious composition. It has provision for storage and immediate recall of sounds devised. Compare the elaborate patch panels with that of the VCS3 synthesiser.

not write this down in conventional musical notation, although it can be written in terms of instrument settings. It was never possible to write the composition of, say, a trumpet sound on a conventional music score, but one can write this electronically in terms of the values of the constituents; so in this way, the actual nature of a sound can be put in black and white so that someone else can recreate the same sound and this has never before been possible.

Clearly the facilities of a synthesiser relate to its use. An amateur experimenter may not need the many duplicated oscillators, noise sources, amplifiers and treatments which are called for by a professional musician or composer. So it is useful that simple modules can be made or bought and added to as required. One must have at least one good tape recorder for the weakness of all synthesisers of low cost is that they have no storage facilities; all work involves short-term events and constant rerecording.

Most if not all modern synthesisers are voltage controlled; tha: is, the conduction properties of the transistors are set by applied d.c. voltages; in this way, several advantages follow at once, the principal ones being linearity of signal with applied voltage change; low impedance of control circuits; and complete absence of hum pick up because one does not trail base (or grid) circuits out from the main apparatus. Hundreds of feet of signal control cable can in this way be run out.

Voltage control is also convenient for a keyboard, since at no time would the voltage across any part of the system exceed about 30V. By controlling the frequency







# **ORGAN KITS**

Build your own Electronic Organ, if you want the best value for your money. Yes you really save over 50% and get the best and up-to-date designs. There are four models to choose from. 

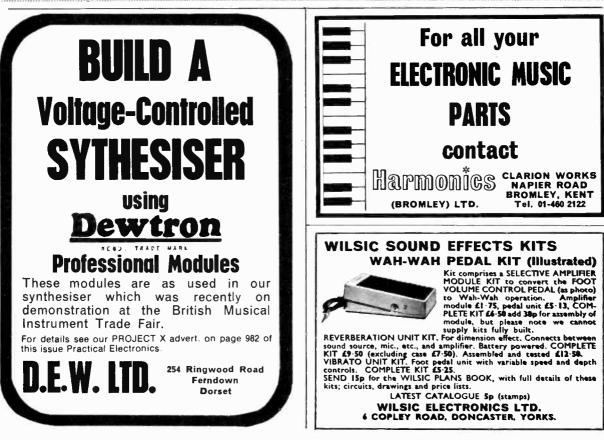
- Portable-4 octave keyboard with 10 voices, 3 pitches-vibrato, at £103.00, P/P £1.50.
- ★ Console—5 octave keyboard with 10 voices, 3 pitches. Keyboard can be split into solo and accompaniment. Vibrato built in amplifier and 50 watt 12" Goodmans speaker at £167 00, P/P £5.00.
- Console—2 x 4 octave keyboards and 13 note pedal board, 29 voices, Vibrato, Delay Vibrato, Sustain Reverberation, Precussion, Wah Wah, etc. at £406 00. Carr. paid on completé kit U.K. only.
- carr. paid on complete kit U.K. only.

We regret H.P. facilities are not available, but components can be bought separately. Trade and overseas enquiries welcomed. Send 25p for latest catalogue.

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### ELVINS ELECTRONIC MUSICAL INSTRUMENTS 8, PUTNEY BRIDGE RD., LONDON S.W.18 TEL: 01-870 4949

Bus numbers 37, 220, 225, 168 pass the door. Nearest stations: Wandsworth Town, East Putney, Putney, Putney Bridge.



Za Green and Andreas and Antion and

of an oscillator by adjustment of the base voltage from a resistor chain contacted by playing kevs, intervals other than those of the equally tempered scale can be produced. This is very useful for exploring the possibilities of the quarter or tenth tone scale—or other subdivisions. This is an approach to the continuous or gliding scale, available on certain instruments like the violin or trombone, but only to a very limited extent; there is no limit to electronic glissando.

Synthesisers have envelope controls; the overall shape of a waveform is called its envelope. By altering the rate at which the wave starts, or stops, or both, profound changes in the character of the sound emerge. One could thus determine, for instance, if it would be a good thing to make the attack of a 'cello longer or shorter, by simply feeding the pick up on the

00000000

'cello into the envelope shaper circuit. These methods have been used on the continent to try to improve the properties and characteristics of some instruments of the orchestra.

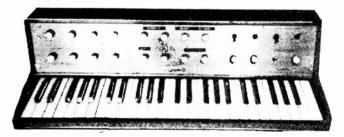
Then there are noise generators. Electronic noise is very controllable and can be useful to augment the impact of certain musical effects. It is also valuable to imitate storm, rain or wind and for this latter purpose can be coloured by a tone superimposed. rather like the sound of wind in telegraph wires. In short, if the synthesiser has enough parameters, almost any musical or abstract array of waveforms can be achieved and recorded: and there are commercial examples of disc records based on purely electronic sound.

Today there are many synthesisers on the market; most have similar parameter facilities, but some are much easier to patch or

cross-connect than others. As experience was gained in their use, controls became simpler and more accurate. The one time room full of gear became a table-top unit as we can see in the illustration of the EMS portable synthesiser.

The ultimate in versatility is the Synthi 100, as used by the BBC and other broadcasting authorities, and made by EMS of London. It has storage facilities and quite elaborate compositions can be realised and performed on it. There is provision to call up anything previously recorded, erase or add to it, and play it at any time. In short, it has all the facilities one could reasonably demand.

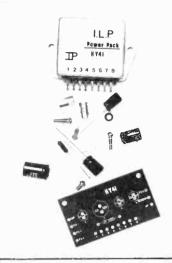
Supplement cover picture by courtesy of St. Giles Music Centre.



The Dewtron synthesiser which combines a whole range of effects in a simplified unit.

The VCS3 electronic music synthesiser produced by EMS (London) Ltd. This small scale voltage controlled studio is capable of producing a great many sound effects by treatment of three oscillatory sources. Signal sources and treatments are labelled down the left-hand side of the patch board. These may be connected in any permutation to the signal input and control input listed along the top by simple jack plugs. Effects produced can then be applied to the keyboard if required.

# Electronics) Ltd



#### THE HY41

The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically 0.05% at 1KHz into 8 ohms!) and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 retains all of the quality characteristics found in with this important improvement the first relation of the advance Mountings and comprehensive the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete month. power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi-Fi but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier.

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.

OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts

R.M.S. continuous. LOAD IMPEDANCE: 4-16 ohms. INPUT IMPEDANCE: 30K ohms at 1KHz. VOLTAGE GAIN: 30db at 1KHz

TOTAL HARMONIC DISTORTION: less than 0.15% (typical 0.05%)

at 1KHz FREQUENCY RESPONSE: 5Hz-50KHz ± 1db. SUPPLY VOLTAGE: + 22.5volts D.C. SUPPLY CURRENT: 0.8 amps maximum.

PRICE: inc. comprehensive manual, P.C. board, five extra components and P. & P.I-MONO: £4.90 STEREO: £9.80

#### UNIQUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier to contain all feedback and equalization networks within an integrated pre-amplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use. Internally the HY5 provides equalization for almost every conceivable input, the

desired function is achieved by use of a multi-way switch or by direct interconnection, Two distinctive features of the HY5 are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from 16-25 Volts and a balance circuit

which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier Specifically and critically designed to meet exacting Hi-Fi standards, the HY5

combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

#### INPUTS

Magnetic Pick-up (within ±1db RIAA curve)  $2mV_{-}47K_{\Omega}$ Tape Replay (external components to suit

head). 4mV. 47KΩ Microphone (flat) 10mV, 47KΩ

Ceramic Pick-up (equalized and compensatable) 20-2000mV. variable. Tuner (flat) 250mV. 100K  $\Omega$ Auxiliary 1 250mV. 47K  $\Omega$ Auxiliary 2 2–20mV. 100K  $\Omega$ 

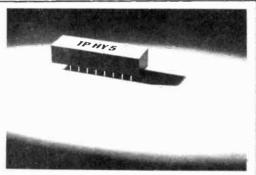
OUTPUTS

Main Pre-amp output 500mV. Direct tape output 120mV

ACTIVE TONE CONTROLS (Bexendall) Treble + 12d Bass + 12db. 12db.

INTERNAL STABILIZATION Enables the HY5 to share an unregulated supply with the Power Amplifier. SUPPLY VOLTAGE

16-25 volts MONO: £3.60



SUPPLY CURBENT 6mA approx OVERLOAD CAPABILITY better than 26db on most sensitive input infinite on tuner and auxl. OUTPUT NOISE VOLTAGE: 0.5mV

POWER SUPPLY PSU45 The versatile P.S.U.45 is designed to supply your HY41's +HY5's in stereo or mono format.

STEREO: £7.20

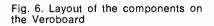
Specification

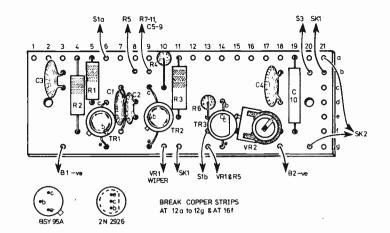
Input: 200-240 Volts. Output: + 22.5 Volts at 2 amps. Overall Dimensions: L. 7"; D. 3.8"; H. 3.1"

PRICE: £4.50 inc. P. & P.



#### **CANTERBURY 63218**





82 is read off the scale, this multiplied by 100 to give 8200, which is the value in ohms of the resistor under test.

#### MARKING THE SCALE

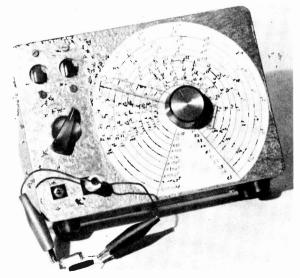
The procedure for calibrating the unit is as follows. With the unit switched to resistance, a resistor of the same value as that switched in by S2 should be connected across the test leads. The control knob of VR1 should be swung towards the high value of the scale until the bridge balances, and then the scale should be marked 100 at this point.

A resistor one tenth of the value of that switched in by S2 should then be connected across the test leads. The knob of VR1 should then be swung to the low value end of the scale until the bridge balances, and the scale marked 10 at this point.

A full range of components should then be used to fill in all the numbers of preferred values between these two points. The same procedure should be used for all five resistance ranges.

#### CAPACITANCE RANGES

Suitable scales for the capacitance ranges may then be marked by the same method, except the



The completed C-R Bridge showing the scale fixed to the front panel and calibrated

components required to mark the ends of the scales will have to be ten times the value, and the same value as that switched in by S2.

The accuracy of the unit will depend almost entirely on the tolerance of the components used in its calibration. The only other factors affecting this, being, the stability of the components used in the bridge, and the care taken when marking the scales.

Most potentiometers have fairly large tolerances, and it is possible that this could cause the unit to be unable to cover the full ranges quoted. If this should occur, it may be corrected by reducing the value of R5 to 910 ohms, or if full coverage is still not obtained, 820 ohms.

#### USING THE BRIDGE

Adjustment of VR2 has to be made before the unit may be used. The range switch should be switched to resistance Range 1, and the earphone plugged in. VR1 should then be adjusted for maximum output in the earphone. VR2 is then adjusted so that the volume is fairly loud, but not excessively so.

To measure the value of a component it is merely connected to the two crocodile clips, and then the bridge switched to read resistance or capacitance as appropriate. If the component under test is wired into a circuit, one of its leads must be disconnected, to prevent other components in the circuit from affecting the reading.

If the value of the component is known, and it is merely to be checked, then the bridge should be switched to the appropriate range, and then balanced (VR1 set for minimum output in the earphone).

#### COMPONENTS OF UNKNOWN VALUE

If a component of an unknown value is to be measured, then the bridge should be switched to a middle range such as Range 3, and then balanced if possible.

Should the bridge balance with the pointer of VR1 on the scale, then a reading may be taken. If the bridge will not balance and it is a capacitor under test, then a lower range should be tried.

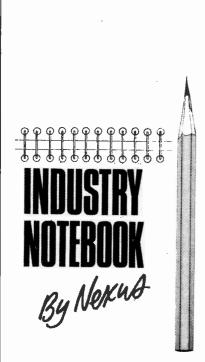
Should this occur when a resistor is under test, then a higher range should be tried. If the bridge balances, but is off the scale, then either a higher range of capacitance, or a lower range of resistance is required.  $\bigstar$ 

#### THE BIG INDUSTRIAL "TRANSFER" MARKET

THE past few weeks have seen a spate of activity in takeovers and mergers.

As forecast in this column in the July issue, Racal were in fact in negotiation at the time and the result was the acquisition of the communications division of S. G. Brown Ltd. from the Hawker Sidde-ley Group for a reported cash sum of about £1 million.

Unitech Group have acquired APT Electronics from the Bonochord Group. APT's range of



power supplies will now complement those already manufactured by the Unitech subsidiary, Coutant, although I understand that APT and Coutant will continue to operate as separate companies. Having hived off APT, the Bonochord Group then acquired Livingston Hire and its subsidiary companies for a sum in excess of £1 million.

#### RATIONALISATION

The old-established Dubilier, itself only acquired a few months ago by a consortium headed by Hambro's Bank and Slater Walker Securities, has now acquired Greenpar Engineering and Kenneth E. Beswick. Greenpar was acquired by Thorn Group not all that number of years back—was it two or three?—and Beswick, who has been in the fuse business for fifty years was a subsidiary of Tremletts. Then there was the merger of Gresham Recording Heads with Information Magnetics Corporation of California, a company little known in the U.K. but wellesteemed in the U.S.A. even though it has been in business only since 1969 manufacturing disc recording heads in competition with IBM and other U.S. giants.

All this flurry of activity (and the examples I have given are by no means all) shows a general trend towards re-grouping in the face of increasing competition and the need to decide which market sector is appropriate to the business you think you are in.

#### CALCULATED RISK

Advance Electronics has gone the whole hog in the competitive electronic calculator market by following up the Executive 8, announced at this year's I.E.A. exhibition, with a further five models including a pocket machine which sells at £52.50 for which Advance expect sales of at least 25,000.

It has been generally thought that the Japanese had the bulk of the world market for electronic calculators nicely sewn up. Not so. See what has happened in the United States when some of the more enterprising manufacturers decided to take the plunge in the low-price consumer field. They are choc-a-bloc with orders. So much so that they are now in trouble.

Computer Design Corporation, one of the entrepreneurs, has been complaining that the semiconductor manufacturers have let them down on the supply of the LSI MOS chips which are the heart of the machines. Computer Design has been reported as having over 4,000 machines on the shelf waiting for components. Computer Design has three suppliers of LSI memory circuits to be on the safe side but delivery dates have been seriously under-estimated, in some cases it is reported the delay is six months.

Another U.S. manufacturer of calculators is complaining not only of late delivery of circuits but also that rejects were as high as 70 per cent although the figure has now improved to 10 per cent.

Nobody has extolled the virtues of LSI more than the semiconductor manufacturers. Now they find they can't cope with the demand. The supply of LSI chips is in the hands of a very few multinational manufacturers who can dictate where available supplies should go. Such is their power that they could make or break a calculator manufacturer if they so wished.

#### ELECTRO-OPTICS

In our September issue I wrote that nobody has yet made a fortune out of lasers, but they could

become big business. I now record that the enterprising Munich Trade Fair Company is promoting a Laser '73 exhibition in their city next September with the forecast that this no-longer novel technology (the laser was first demonstrated in 1960 by Theodore H. Maiman in Hughes Research Laboratories) will have a civil market alone worth billions of dollars. Be that as it may, the exhibition should itself prove a profitable affair because despite its title it will cover the whole field of electro-optics which is about the fastest growing sector of electronics.

#### . . . AND MICROWAVES

If electro-optics is a booming industry so is microwaves judging from the response to the *Microwave* '73 international exhibition and conference scheduled to be held at Brighton next June. Stand area has just been increased by 40 per cent to cope with the demand for space. So far, nearly 80 companies from U.S.A. and Europe have booked space as exhibitors and the accompanying conference, I understand, will be heavily slanted towards applications of microwave technology rather than theory.

#### "FAX" MARKET OPENS UP

Facsimile transmission was first demonstrated—at least in principle—by Alexander Bain, a Scottish clockmaker, way back in 1842. Now, 130 years later it is really beginning to be big business and the high-powered companies are already moving in.

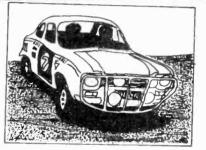
Plessey is one of the recent entrants with an American designed machine built in Japan. Plessey did a test marketing operation last year before deciding to go ahead. The operation is already swinging and Plessey expect to be selling over 5,000 a year by 1976.

Most unusual use for the Plessey "Remotecopier" was in "British Steel" during the recent singlehanded transatlantic yacht race when skipper Brian Cooke had one aboard for reception of weather charts from which he was able to predict wind changes several hours ahead of his competitors. But the big market is in offices where it is being promoted as "business tool".

But, as the market opens up, so the competition increases. There are seven major manufacturers in the field. In the U.K., latest entry is EMI who is marketing EMIfax machines through its subsidiary company SE Computer Peripherals. The machine is produced by the German company Rudolph Hell which happens to be a subsidiary of Siemens.

# TWO SPECIAL CONSTRUCTIONAL PROJECTS FOR THE PARTY SEASON

# RALIY DRIVING GAME



Five different "cars" participate in a competition of "driving" skill. Which driver can keep his car on the road, under optimum acceleration and braking conditions, depends on the acquired skill of simulating his driving conditions on a control panel, bearing in mind the type of car he is driving.



# HOOTING

Leave the hooting owl in a conspicuous place in your lounge and someone is bound to pick it up to take a closer

look. He arouses curiosity, he is a pet for the children; he just likes to sit and gaze into space. He feeds on small batteries (about twice a year when contented). But if he is frequently handled, his appetite goes up to help him hoot.



DECEMBER ISSUE ON SALE NOVEMBER 10



### ENTRY REGISTER LOGIC AND HARDWARE

LAST month we dealt with the construction of the keyboard logic panel which has the function of converting each key depression into suitable logic signals for entry and control of data.

As stated in the first part, entries of up to six digits can be made. The subject of this month's part is the Entry register and associated logic which has the function of storing these six entered digits in their correct order for use in calculation to come. A six digit number can also be placed in a memory for use at any number of times during the calculation.

This part also discusses the possibility of a "fixed constant" key whereby a single depression of an extra key causes a chosen number (such as  $\pi$ ) to be entered into the entry register.

#### ENTRY REGISTER LOGIC

A block diagram of the ENTRY register complex is shown in Fig. 5.1, which is an expansion of the ENTRY register section of the overall arithmetic section block diagram from Part 1 (Fig. 1.3).

The operation of the ENTRY register and its associated constant store and display multiplexer is relatively straightforward, each of the possible six decimal entry figures being stored as a four-bit parallel B.C.D. code, requiring a total of  $6 \times 4$  (= 24) separate storage bistables.

The 24 bistables or flip-flops are arranged as a shift register six decimal digits long and four B.C.D. digits wide, so that with each clock pulse a complete B.C.D. group (i.e. one decimal digit) is shifted into, or down, the register.

Each time a number key is pressed, the keyboard circuits (which were described last month) staticise the corresponding B.C.D. code and generate a single

clock pulse, which is used to clock the four bit code into the ENTRY register.

Pressing a second or subsequent number key causes the first group to shift down the register to the "left," its place being taken by the new entry. This process can continue until the first entered number ends up in the extreme left-hand location of the register, after which any further entries will cause that first number to be destroyed by being shifted out of the end of the register.

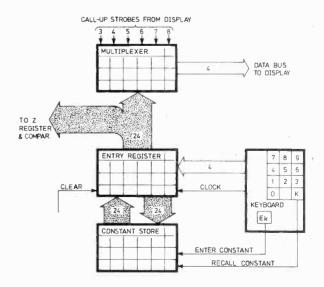


Fig. 5.1. Block diagram of the Entry Register complex showing the inputs and outputs

Note that the ENTRY register accepts the most significant digit of the complete figure entry *first*, just as we write figures ourselves. This makes the ENTRY register a "left-shift" register and distinguishes it from the other registers which normally accept the least significant digit (L.S.D.) first, and are thus termed "right-shift" registers.

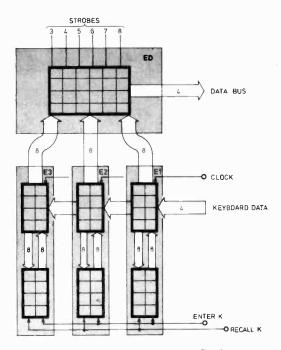
#### CIRCUIT PARTITIONING

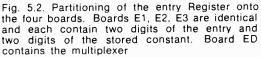
The ENTRY register and associated constant store and multiplexer, are built on three separate plug-in cards. The way in which the circuit is divided among the four cards is shown in Fig. 5.2., where it is immediately apparent that three of the cards are identical.

In effect partitioning the circuit in this way results in three independent shift-registers, each two decimal digits long and four B.C.D. digits wide, and each having a proportionate amount of constant storage.

In use these small units are connected in series, via the edge connector wiring, but because each card is identical they can be freely interchanged and this is a great help in tracing any faults which may occur.

The display multiplexer is built on a separate card with a 44-way edge contact instead of the 22way type used for the other three cards, and although not shown on the block diagram, this card also houses two SN7440 buffer gate packages used to drive the CLOCK and CLEAR lines of the register.





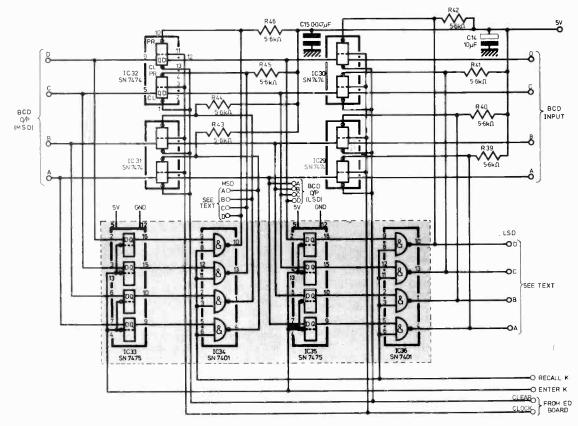


Fig. 5.3. Circuit of the Entry Register board E1. Boards E2 and E3 are identical to this. The dotted line encloses the constant store which may be omitted if not needed

#### **REGISTER CIRCUIT**

The full circuit diagram of an individual register board is shown in Fig. 5.3, three of these sections being required to form the complete register.

The circuit splits horizontally into two logical sections, the upper four i.c.s forming the shift register proper, and the lower four forming the constant store.

The components forming the constant store are enclosed in a dotted box for easy identification, and it is these i.c.s which may be left out if required, along with their associated wiring.

The eight bistables which comprise the shift register itself are SN7474 dual D-type elements, chosen because of their flexibility and low power dissipation.

Both of the two bistables in each package have independent clock, data, preset and clear inputs, together with true and inverted outputs. D-type flip-flops are more suitable than the J-K types for shift register application because they only require a single data (D) input which has the effect of reducing wire interconnections and leaving pins available for the independent preset function without recourse to the more expensive 16-pin package.

To form the register the flip-flops are connected in series pairs, there being four such pairs on each board to cater for the four separate digits (A, B, C, D) of the B.C.D. code.

Each vertical group will contain four binary digits which, taken together, represent one of the decimal numbers zero to nine.

The clock input to the register has to drive all of the flip-flop clock inputs in parallel so that after each pulse the data in a particular four bit group move one place to the left, and are themselves replaced by new data.

It is necessary to clear all the data from the ENTRY register simultaneously when required by the programme, and to this end all the CLEAR inputs are connected together to a common input, a low level, or "ground" condition on this input will set the Q outputs of all the flip-flops to zero, and the  $\overline{Q}$  outputs to one.

#### CLEARING OPERATION

The contents of the ENTRY register are transferred in parallel to the z register early in the arithmetic programme, except during MULTIPLY, when they remain to be compared with the contents of the counter.

When the register contents are finished with the ENTRY register is cleared by a signal from the programme, ready for subsequent entries to be made by the operator.

Any errors made during figure entry (e.g. pressing "8" instead of "6") can be corrected by pressing the CLEAR ENTRY key which also has the effect of clearing the contents of the register ready for new data.

#### CONSTANT STORE

The constant store is shown beneath the ENTRY register, and is of the same 24 position capacity, so that any entry made can be stored for further use in a calculation.

This store is not essential to the correct operation of Digi-Cal and can be left out permanently or temporarily if desired, without the need for any modification to the rest of the circuit. In operation, with the desired constant entered in the usual way into the ENTRY register, the ENTER K (EK) key is pressed, which duplicates the contents of the ENTRY register in the constant store by means of a single, parallel, 24 bit transfer.

Constants stored in this way remain available until they are either replaced by a new number or the machine is switched off, there being no requirement or provision for clearing this store, other than by entering a constant which consists of all zeros.

Constant recall is carried out by pressing the  $\kappa$  key in the numeral section of the keyboard whereupon a reverse transfer occurs, in parallel, from the constant store to the ENTRY register. This operation does not destroy the contents of the store so that any constant can be used as many times as required in a calculation.

The constant store need not be used simply for storing constants in the accepted sense, since it will act as a memory of any result or intermediate answer if required, provided these numbers are entered through the keyboard in the usual way.

This mode of operation really acts as a substitute for pencil and paper, and can be most useful at times in long calculations.

#### DISPLAY MULTIPLEXER

The display board, described in Part 3, will display either entries or answers depending on a signal from the programme, entry and answer data being routed to the display via two four-line "buses."

Timing signals, in the shape of "character call-up" strobes are produced by the display board to enable the four line buses to carry all the data in their associated register to the display in a time-shared sequence, the sequence being produced in a multiplexer circuit.

There are two multiplexers in Digi-Cal, one for the eight digit answer and the other, which we are interested in here, to handle the six digit entries.

Only strobes three to eight inclusive are used by the ENTRY multiplexer, each of these allowing only one four-bit B.C.D. digit on to the bus at any instant in time, the strobe direction being from most to least significant digit (M.S.D. to L.S.D.).

#### CONSTANT CIRCUIT

Simpler storage elements can be used for the constant store since there is no requirement for PRESET or CLEAR inputs, and the devices chosen to fulfil this function are the SN7475 four bit latches.

Like the SN7474s these quad latches require only single "D" inputs, the data to be stored being of course the outputs from the shift register flip-flops.

Clocking is controlled by the ENTER K key, via a buffer gate mounted off the board. There is no need to "debounce" the output of this switch because of the simple "gated latch" operation of the SN7475 flip-flops and the static nature of the inputs during the entry operation.

Stored constants are returned to the shift register via the SN7474 PRESET inputs, transfer being controlled by the SN7401 quad NAND gates, which have a common input enabled by the RECALL  $\kappa$  key, via a buffer gate. The PRESET inputs of the SN7474s are "active low," i.e. they set the Q output to "1" when an "0" input is present, and so require inverted data from the constant store. This inversion is provided by the SN7401 gates.

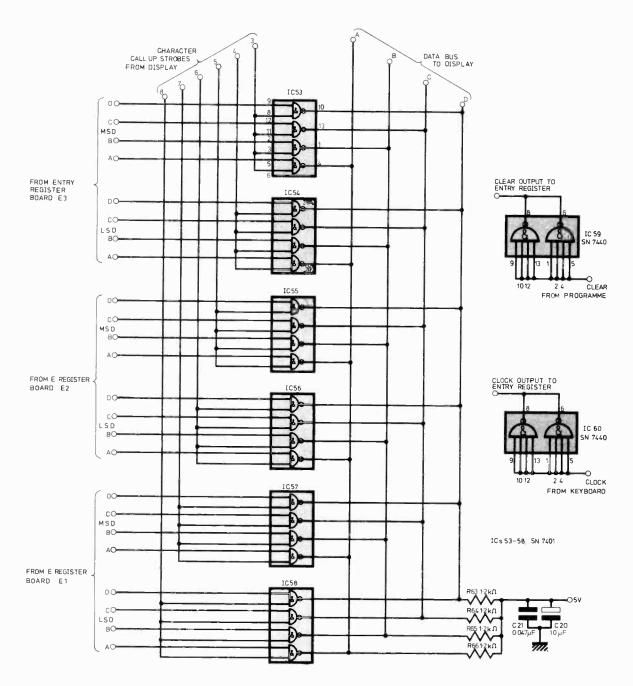


Fig. 5.4. Circuit of the Entry Register Board ED. The two SN7440 i.c.s are simply buffers and only placed on this board for convenience

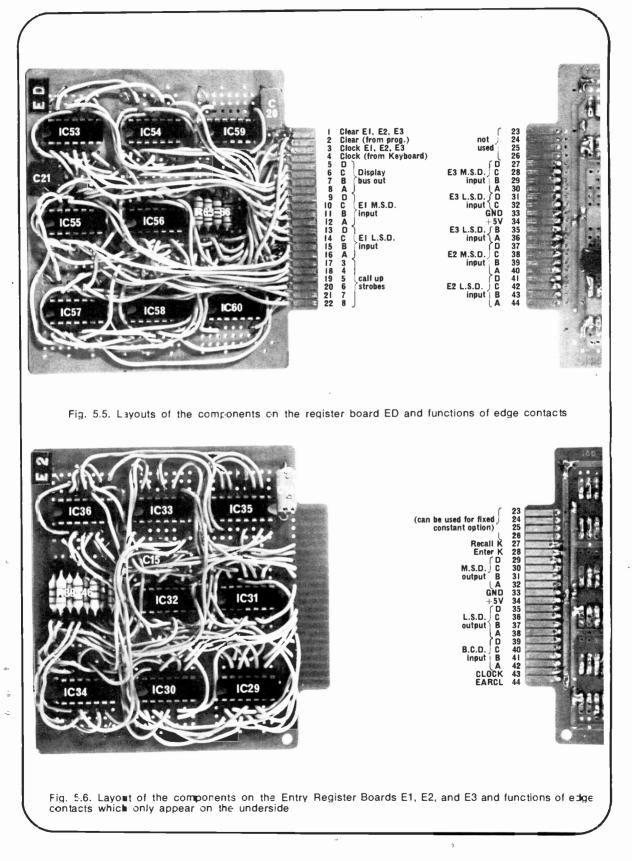
#### MULTIPLEXER CIRCUIT

The circuit of the display multiplexer, Fig. 5.4, is quite straightforward, it being formed only from gates and resistors. Each SN7401 gate package has its four inputs wired to the corresponding four B.C.D. output pins on the appropriate register board, there being a separate gate i.e. for each decimal digit.

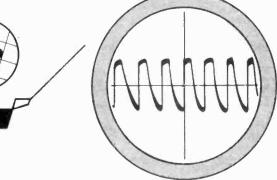
The common input to the four gates in each i.e. is driven by the "character call-up" strobe appropriate to that digit. All the "A" outputs from the gate packages are wired together, as are the B, C, and D outputs, to form the four line display bus, which is referenced to the 5V line by the four 1.2 kilohm resistors.

Note that SN7401 open collector gates are essential for this circuit, SN7400 gates being unsuitable due to the fact that several gate outputs are connected directly together to provide the "Wired or" function.

Interconnection of outputs is not permitted with the basic TTL gate because of the "active pull-up"



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### COMPONENTS . . .

REGISTER BOARDS	E1,	E2,	E3
-----------------	-----	-----	----

**R39-R62** 5.6k $\Omega$  + 10%  $\pm$  W carbon (24 off)

#### Capacitors

Resistors

C14, C16, C18 10μF 15V elect. (3 off) C15, C17, C19 0.047μF (3 off)

#### **Integrated Circuits**

IC29-IC32, IC37-IC40, IC45-IC48	SN7474 (12 off)
IC33, 35, 41, 43, 49, 51	SN7475 (6 off)
IC34, 36, 42, 44, 50, 52	SN7401 (6 off)

#### Printed circuit boards and sockets

Type DL109/22 (3 off) (Available from Shirehall Electronics Ltd., Station Yd., Borough Gn., Sevenoaks, Kent) SK2-4 DPK165 edge connectors (Shirehall) (3 off)

output stage. The SN7401 is a gate specially produced to allow "Wired OR" function in TTL systems, and has no pull-up device incorporated.

The two SN7440 buffer gates are positioned on the multiplexer board but have no direct connection with the multiplexer circuit. The two buffers in each package are wired in parallel to give sufficient drive capability (or "fan-out") to handle the large load represented by the ENTRY register CLOCK and CLEAR lines (48 loads and 72 loads respectively).

#### CONSTRUCTION

Wiring up is carried out in the same fashion as the display board described last month, though the plug-in cards are much easier to work with because of the pre-tinned finish and the ready formed i.c. pads which obviate the need for "spot face" cuts. (Note that the three boards are identical, each board taking one third of each group of components shown above)

#### **REGISTER BOARD ED**

**Resistors** R63-R66 1.2k $\Omega \oplus 10\% \frac{1}{2}$ W carbon (4 off)

#### **Capacitors** C20 10μF 15V elect. C21 0·047μF

Integrated Circuits IC53-IC58 SN7401 (6 off) IC59, IC60 SN7440 (2 off)

Printed circuit board and socket Type DL109/44 (Shirehall) SK1 DPK165 edge connector (Shirehall)

By referring to the appropriate circuit diagram and the i.c. layouts shown in Fig. 5.5 and Fig. 5.6 wiring is quite straightforward, the only points to remember being the need for links from the power buses to each i.c. and from these power buses to the appropriate edge contacts, along with the need to orientate the i.c.s correctly before soldering into circuit.

All wiring up is carried out on the blank side of the boards, using single core wire.

The best order to complete the construction of the boards is as follows: ED board (display multiplexer) including the buffer gates, on a Shirehall DL109/44; then the register section (SN7474s) of boards E1, 2, 3, on Shirehall DL109/22 cards; and finally when the above sections are operating correctly, the constant store and if required, the fixed constant option, may be added.

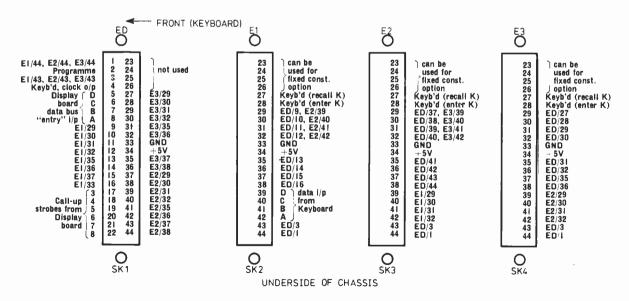


Fig. 5.7. Interwiring of the four entry register boards is carried out using the edge connectors fitted in the large hole cut in the chassis plate. The edge connectors are fitted so that the boards plug in from above. Destinations of wires are shown next to each contact (e.g. contact 30 on socket E1 should be wired to contact 10, board ED, and contact 40, board E2)

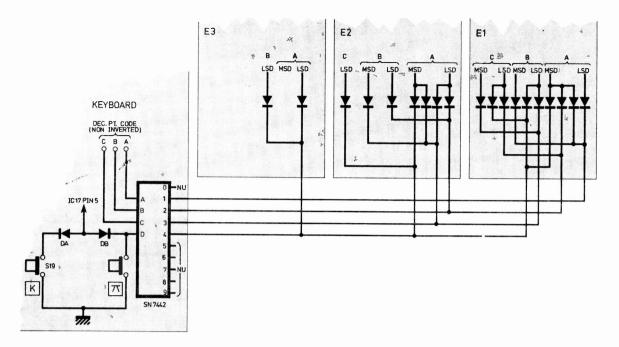


Fig. 5.8. Extra wiring to allow for a fixed constant key in addition to the K key. The diodes shown allow for the number  $\pi$  or a close approximation, to be placed in the entry register when a single key is depressed. The SN7442 is an extra i.c., space for which can be found on the Keyboard panel

#### INTERCONNECTION AND TESTING

Under chassis edge connector wiring can be started early in the construction sequence to allow operational testing to be carried out as necessary, and this task can be readily completed by referring to the edge connector layout and wiring tables (Fig. 5.7).

The tables are easier to use than wiring diagrams

their instructions being implemented by simply wiring the edge connector pin required to all (or any part as required) of the destination listed.

When the register section of one or more of the boards is completed, keyboard entries should be possible by pressing a sequence of number keys, provided of course that the display and keyboard are already functioning, and that the required CLOCK, CLEAR, and data interconnections have been made.

Dec. nt	Required	Decimal digit identification red Board E3 Board E2 Board E3							
Dec. pt setting	"π" value		M.S.D.			L.S.D.	Boar M.S.D.		B.C.D. digit
1	3.1	{					×	×	D C B A
2	3.14	{				×××	×	×	D C B A
3	3-142	{	·		××	×	×	×	D C B A
. 4	3.1416	{		×××	x	×	×	××	D C B A

Table 5.1 EXAMPLE OF FIXED CONSTANT KEY DIODE PLACING

The only temporary test connection required to render the circuit operational is a ground connection to the CLEAR buffer gate input to prevent a permanent CLEAR signal being produced.

#### FIXED CONSTANT OPTION

SN7401 open collector gates are used in the RECALL K path in preference to SN7400 gates because they can be used to perform the "Wired or" function at their outputs. In practical terms this means that it is possible to "get at" the SN7474 PRESET inputs for use with other data sources than the constant register without altering existing logic.

This possibility opens the door to all sorts of custom modifications to enhance the usefulness of Digi-Cal, and to show how simple some improvements can be, let us consider a scheme for entering a constant which is very commonly used, without taking up space in the constant store itself.

The constant is the ubiquitous Pi  $(\pi)$ , but in fact any constant which is likely to be often used is suitable for our purposes, the object being to have the chosen constant always available at the press of its own *separate* key, without interfering with the operation of the constant store which remains available for routine use.

This modification requires only the incorporation of a number of diodes into the ENTRY register boards and the addition of an extra "PI" key on the keyboard with a single extra i.c. The circuit for such a scheme is shown in Fig. 5.8.

An extra diode must be connected to the keyboard logic panel (Fig. 4.2), between the top of the "K" switch S19 and pin 5 on IC17, the previous direct connection being broken. This diode, marked DA in Fig. 5.8, together with diode DB forms an or gate so that pressing either key "K" or " $\pi$ " causes the automatic normalisation to be inhibited.

#### **OPERATION OF CONSTANT FACILITY**

The principle of operation is that when the "PI" key is pressed one of the outputs of the SN7442 decoder will go low, depending on the setting of the decimal point switch.

The energised output is used to PRESET a number into the ENTRY register via diodes wired in wherever a "1" is required. The diodes are connected to the PRESET inputs at the points marked in Fig. 5.3, and perform the "Wired OR" function with the SN7401 outputs.

The whole circuit acts as a diode "Read Only Memory," where the contents of the memory are programmed at the construction stage by wiring in diodes where required.

The memory contains four separate numbers to allow for the four possible decimal place selections. Table 5.1 shows how the placing of the diodes is determined, a cross marking the position of each.

This circuit can be very useful if Digi-Cal is to be used for calculations containing an often used constant, and can be substituted for the constant store or used to complement it. If this option is never likely to be taken up it would be possible to use SN7400 gates in the RECALL path instead of the SN7401s and resistors, but note that the pin connections for the SN7400s are different.

### Next month: Logic and construction of A and Z registers.

# **NEWS BRIEFS**

#### New Data Transmission Technique

A NEW technique of data transmission is to be tested in a forthcoming experiment involving the Post Office and computer manufacturers and users. The system is known as "packet switching" and is basically the transmission of computer data in self-contained, addressed blocks like a series of high speed telegrams.

The user of such a system sends out the data together with the address of its destination and the data is automatically routed to that destination by the system. The need to set up a direct link between sender and receiver before transmission is thus eliminated.

Circuits connecting packet switching exchanges can be used for carrying packets sent by other customers in the time intervals between packets in a series making up a complete message. Because many signals travel on the same wire a large number of low capacity connections to a multi-access computer can be replaced by a single high capacity connection.

The system has the advantage of lower error rates than conventional systems and enables two terminals with different data transmission rates to be connected.

#### **New Range of Calculators**

A BANGE of five new electronic calculators has just been announced by the British firm Advance Electronics Ltd.

Because of large scale integration whereby all the electronics can be placed in a handful of discrete packages, assembly costs of complex instruments have been reduced drastically. The features which Advance are promoting are reliability and value for money rather than such dubious advantages as extreme miniaturisation.

The new range consists of four desk-top calculators known as the Executive 16 range and a pocket-sized calculator known as the Executive.

The Executive 16 range all feature 16 digit capability and a keyboard specially designed for high speed work. The keyboard allows a key to be pressed even when another has already been pressed providing that they are released in the correct order.

The basic model (retailing at £95) has the four basic arithmetic operations as well as a "%" key.

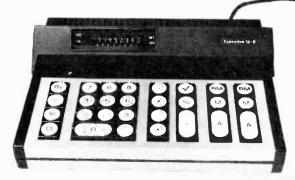
The 16 + 1 (at £115) has a versatile memory with full 16 digit capacity which can also function as an accumulator for automatic list totalling.

The 16 + 2 (at £145) is specially suitable for V.A.T. calculations featuring two memories.

The 16R (at  $\pm 175$ ) has all the features of the 16 + 2 but also includes a square root key.

The fifth member of the range is an eight-digit portable (at  $\pm 52.50$ ) with a plinth for desk use and mains power supply option extras.

Shown here is the Executive 16R which is the most versatile of the Executive 16 range





### AUDIO FREQUENCY ANALYSIS

THE analysis of audio frequencies by separating a selected frequency component from a complex signal can be useful for any number of purposes. In BP 1 282 487 The Standard-Triumph Motor Company Limited describe an analysis system which they have found particularly suitable for checking the audible noise from a gear box to isolate gears which are unduly noisy.

Past practice, directed at separating the noise of the gear in question from the overall gear noise, has been to compute the main frequency of sound produced by any two gear wheels by considering their teeth number and speed. With the main frequency component thus known it is isolated by applying a reference signal (from a tone wheel) to "beat" with the selected frequency.

A simple bridge and capacitor arrangement is used to measure the amplitude. But the snag has been poor low frequency response and confusion due to ripple in the output waveform due to the presence of rectified a.c.

In the new invention a tone wheel produces an audio frequency signal which is fed to the tone input. The noise signal to be analysed is applied to the noise input and the tone and noise signals are respectively amplified and applied to three transformers which constitute a balanced modulator.

The secondary of one transformer is applied to the primary of another via a ring modulator. The secondary of the latter is fed to a 100Hz low-pass filter which passes the beat signal only and blocks all other residual signals (reference and noise).

The filter output is amplified, rectified and fed to a voltage comparator. Two gate drive transistors control a gate, which is opened once every half cycle, allowing a storage capacitor to sample the smoothed beat signal voltage available when the gate is open.

The criterion of the invention is the feature whereby the storage capacitor mentioned above is connected after the gate so that it can only sample the smoothed voltage available when the gate is open. Previously the storage capacitor would have been connected across the output of the rectifier and so responded to a low frequency ripple. This will of course confuse the true response.

BP 1 279 134

#### ELECTRONIC CONTROL OF SORTING

N BP 1 279 134 Gunson's Sortex Limited of London E3, describe a fairly simple method of detecting the difference between mixed objects to be sorted, e.g. between scrap polyethylene and scrap copper. They claim that if modified their device can also be used to sort objects made from different types of plastics material.

The basic feature is that the mixed objects falling down a chute pass through a coil (in Fig. 1) with a small number of turns. This coil is part of a tuned circuit connected into an oscillator circuit.

The output of the oscillator is used to control a sorting deflector which pneumatically deflects objects of one type.

The circuit (Fig. 2) shows the coil together with the tuning capacitors and transistor functioning as a Colpitts oscillator. The frequency produced is usually within the range 200 to 300 MHz.

When an object passes through the coil a positive going change occurs in the output signal from the oscillator. The change in output signal is amplified and fed to a magnitude detector stage. This transistor amplifies and inverts to provide a negative-going signal which can be used to drive a power amplifier for controlling the deflector mechanism.

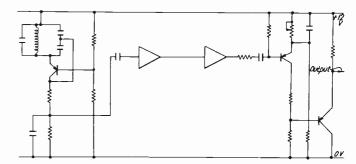
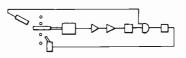
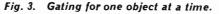
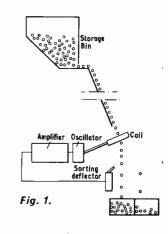


Fig. 2. Basic amplifier and switching circuit

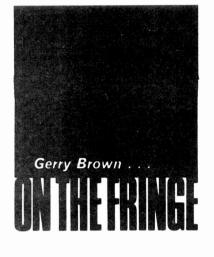
To distinguish between different objects of similar material, the detector transistor switches according to the signal rate of change. Fig. 3 shows an arrangement whereby an "object presence" sensor makes sure that the deflector operates only when required by one object.

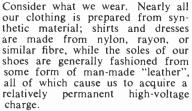






Practical Electronics November 1972





This charge is positive and, what's more, according to some researchers causes most of the fatigue and general debility experienced these days. The question is, should we drag around discharging chains in order to maintain equilibrium, or would we be better off minimising "droop" under a d.c. version of the National Grid?



#### CHARGE

Back in Queen Anne's time, the superstitious held that "air electricity" could have a profound influence on both behaviour and on general disposition. Indeed, such expressions as, "he has the wind in his tail" probably originated this long ago, since it was common practice in European courts to treat felonies with more tolerance whenever the ldes of March were blowing.

common with many old ln beliefs which have subsequently become contemporary facts, air electricity is no exception. Apparently it is ions which are responsible for this effect on people. Ions are atoms (more likely, in this case, molecules) which have either lost or gained a number of electrons; those with extra electrons become negative ions, and those losing electrons, positive ions. For some obscure reason the positive ions make us feel tired and lower our capacity for work, while negative ions produce just the reverse; as when, for example, a thunderstorm has just passed.

Prior to a storm it has been noticed, though, that the air charge is predominantly positive so if the theory is right this would correlate with feeling "lifeless" and "headachey" at such times.

Thunder and lightning aside, in spite of our technical enlightenment we probably do more to make our lives uncomfortable than the "superstitious" people of yesterday.

#### FIRE WITHOUT SMOKE

Since we have had every opportunity to become fully hyperconscious about pollution and its effects, it is staggering to think that most of us are quite uncritical of the quantity of such belched into the atmosphere during that evening



on November the 5th. This is particularly hard to bear when one considers that fireworks have advanced little, technically, for the thousand or so years since the Chinese began using them; albeit, the bangers were likely better then!

So, since this burning sense of fun remains with us to relive the next year, and generally costs more per shower of sparks than it did the year previously, a nagging feeling is left that perhaps the time has now come to rethink the Guy Fawkes night phenomenon. And who could be more fortunate than us to enjoy a relatively luxurious imagination in this respect. While there's still a little time left before this year's "big burn", and before Dad sets light to the poplars again, let's see how realistic a smokeless 5th might be.

For starters, we could do-away with the conventional Catherine wheel and substitute it for a spinning frame of lights, faded electronically in sequence from the centre, and fed via slip rings on the drive shaft. Then add a bit of novelty to roman candles by employing solenoid-released, springejected, miniature mercury cell powered micro-lamps. Or what about capacitor discharge "canon crashers" with extra large xenon flash tubes for realistic back-up?

flash tubes for realistic back-up? Occasional bursts of envelope shaped white noise would not come amiss, particularly in synchronism with periodically operated, brightly lit water fountains. Realistic smoke would be produced with solid CO.

With a philosophy like this fireworks could be arranged to have a virtually indefinite "burning time", and would hardly ever wear out. Honestly, when you consider just how much electrical power is available for the really enthusiastic, it's a wonder someone hasn't done a back-garden Son et Lumière already!

#### EYE-AYE!

Sensory deprivation still remains a bit of an enigma; indeed, the discovery of every new effect has raised an even greater number of questions about its possible mechanism.

Many of the problems associated with experiments up until now could well have been attributed to the rather general nature of the experiments, and further complicated by interaction (or lack of it) between large numbers of unstimulated areas in the brain.

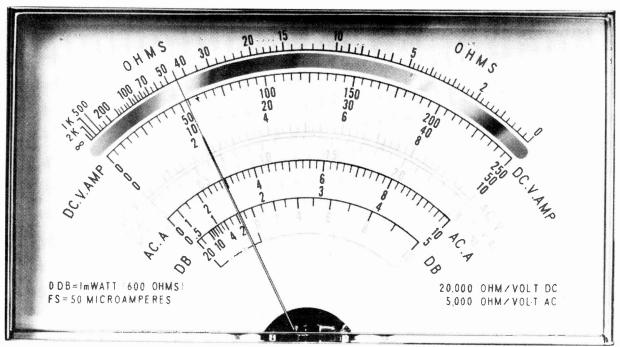
Although the days are not yet past when experimental volunteers need be suspended in tanks of warm water for hours at a time, there does appear to be a modest breakaway toward more systematic examination by limited deprivation of inputs to individual senses.

Employing this technique, two workers at the University of Manitoba have recently discovered an effect very similar to that which occurs following actual severence of sensory nerves. For their experiment each of 15 students were asked to wear a light-blocking patch over one eye during a working day on the campus. The uncovered eye was checked periodically for sensitivity to flickering, rather than continuous light.

On aggregate, uncovered eyes were initially less sensitive, but, given several hours, subsequently began to increase in sensitivity, eventually exceeding even the normal level. In some instances students reported that this hypersensitivity lasted for days afterwards.

These odd happenings could, by implication, be no mere trifle, especially in the context of stimulus withdrawal for any great length of time. Paradoxically, though, an additional effect was that the covered eye maintained normal sensitivity throughout the tests. Perhaps Nelson, bless his heart, really didn't see any ships after all!

Practical Electronics November 1972



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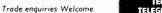
load dimmer. Switchable for response to High, Mid and Low frequency audio input signals, selected by fascia control. D.J. Pulse-Flash push button fitted to fascia for manual flashing of lamp load. Neon load indicator firred

of lamp load. Neon load indicator fitted. Constructed on glassfibre printed circuit material with attractive black anodised fascia panel, Input and load connections via Cinch printed circuit connector. Maximum Load IkW at 250V 50Hz. R.F.I. filtered. Price: £13 each post free U.K.

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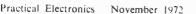
There is a master full-wave triac circuitry, There is a master-level control, together with independent sensitivity controls for each channel. The original minimum ambient light level controls have been redesigned permitting their use as faders; allowing dimming from max. to zero at the turn of a knob. R.F.I. suppression is now incorporated as standard as well as provision for D.J. "Pulse-Flash" controls. The choice of two inputs enables operation from both high and low power amplifiers. Max. power 15kW per channel at 240V a.c. Complete assembly built and tested.

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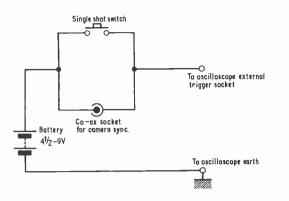


Fig. 1. Final circuit for adding a one shot facility to an oscilloscope



it should be selection of readers' suggested circuits. emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.

payment according to its merits.

SENSITIVE LIGHT SWITCH

This is YOUR page and any idea published will be awarded

ENCLOSE a circuit diagram, Fig. 1, of a very sensitive light operated switch which may be of interest to readers.

When the light dependent resistor PCC1 is illuminated its resistance is low making the inverting input of the i.c. positive with respect to the noninverting (the potentiometer VR1 is adjusted to ensure this) so the output will be at 0V and the transistor TR1 will be cut off. As the light level falls the l.d.r.'s resistance increases until the inverting input is more negative than the non-inverting. This causes the amplifier to saturate and the output rises to +12V, which switches on TR1 and operates the relay.

The point at which the relay operates is controlled by VR1, and is really stable as supply voltage variations have small effect because both voltages vary together (I.d.r. junction and VR1 wiper).

If the reverse operation is required (i.e. relay operates as light increases) connect the l.d.r. to 0V and resistor R1 to  $\pm 12V$ .

> R. S. Girdwood. Norwich.

AVING a double beam oscilloscope which does not have an integral single shot facility, it was felt that single shot operation would be useful in photographing traces. Inspection of the circuit diagram of the scope had shown that with a judicious choice in the setting of the trigger level and stability controls together with the application of external trigger pulses, a very simple system could be used to initiate single sweeps.

The external input to the trigger circuit of my scope (D51) can be a positive or negative going waveform of amplitude greater than three volts. The current required was small, so an old PP3 battery could be used until corrosion rendered it useless.

All that was thus required was a battery and a push on/release off switch connected in series with the external trigger socket of the 'scope.

It was realised that the switch could be replaced by the flash contacts of a camera. Use of a slow shutter speed and X (or electronic flash) synchronisation allows the trace to be recorded on film even under fairly high levels of ambient illumination.

The final circuit is shown in Fig. 1, the flash contacts being wired in parallel with the push on/ release off switch to allow setting up and single shot operation without unnecessary wastage of film.

To set the equipment up, turn the trigger selector to external. Next turn the stability control from the free-run portion of its range until the trace just disappears. Turning the trigger level control from "auto" through the set level area of its range should at some point produce one trace. With the control in this position apply the external trigger pulses and optimise the settings of the stability and trigger level controls for reliable single shot triggering. The 'scope is then set up.

R. A. Shackleford, Glasgow.

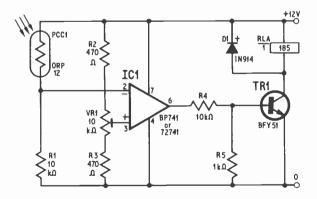


Fig. 1. Circuit diagram for a very sensitive light switch



Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

### **True values**

Sir-Messrs. Bobker, Harrison, and Barrett have a rather strange sense of values (*Readout*—Oct. 1972). They seem to have been They seem to have been caught by the general malaise that is creeping over many sections of society that if you cannot make something for a "profit" then its not worth bothering. They seem to have blinded themselves with the financial matters of the subject and have forgotten that your magazine is not designed with commercial interests in mind but to offer amateur enthusiasts a source of upto-date knowledge on all aspects of electronics coupled with instructions on how to make equipment.

You cater for amateur hobby interests and the value of a hobby beyond price if you is are sufficiently interested in the subject. The price it costs to make something and learn from it is irrelevant --provided one can actually afford the expenditure. If one cannot afford the price of making something then one can still get a high degree of satisfaction out of reading and understanding, in a very practical sense, the workings of a sophisticated piece of equipment.

*Digi-Cal* is the first publicly issued inside story of the workings of a desk calculator and as such is a long waited for event on the amateur front; the three correspondents named above should be thankful that at last someone has gone to the trouble of designing such a piece of equipment and is also capable of explaining its workings in a clear and straightforward manner.

Your correspondents have got the wrong end of the stick. If they want something cheap then they should buy it. If they want real value for money then they should make an effort to scrape together the money to do something worthwhile. Just for a moment let's consider whether they have their commercial facts right. Let's assume they have bought their cheap commercial instrument; something goes wrong with it that is outside the guarantee. What do they do (assuming no inside information regarding *Digi-Cal*)? Of course they have to lay out more money to get the equipment repaired. If they had dropped or crushed their pocket sized instrument it is quite likely that they will have cracked the single i.c. element. How much does this cost to replace?—probably in the order of £24 to £40 depending on the type. How much would it cost if a similar thing happened to *Digi-Cal*? Certainly a lot of time sorting out the fault—but in material costs probably not more than a few pence!

I'm afraid I do not agree with the last paragraph of your editorial. Digi-Cal is NOT "priced out"—if anything it is absurdly cheap. An annual subscription to P.E. costs t2.65; the electronics, mathematical knowledge, and logic experience that can be gained from a series such as this is far beyond the learning one can gain in a whole year of university work in the same field.

My interest is in the field of digital electronics, but I know very little about calculators. This hole in my knowledge is now being filled. I would like to congratulate Mr. Coles for making such an effort in designing a topical, useful and educative project, and I measure that most enthusiastic readers would agree with me.

> Michael J. Hughes, Westerham, Kent.

### **Raudive** voices

Sir—Mr Morton (September, '72) would appear not to have read "Breakthrough" as well as he might have. I quote, "Scientific tests have shown (in a Faraday cage, for example) that these voices originate outside the experimenter and are not subject to auto-suggestion or telepathy." Also, "They are twice the speed of human speech and of a peculiar rhythm which is identical in the 72,000 examples so far examined." Would different experimenters all have the same style of modulation?

There is the possibility that the tape irregularities themselves, when

passed through a high-gain amplifier, might sound like words to anyone who expected (however unconsciously) words to be there. However it sounds unlikely, but, until we know more, it must be an open question.

Michael Fleming, Solihull, Warwicks.

Michael Fleming raises three issues: one trivial and two worth a closer look. I will try to deal with them in turn.

First, the "tape irregularity" explanation, which is in fact a politer way of saying "imagination" or even "fraud". Can we doubt that a phenomenon so seemingly preposterous as the Raudive voices would have been examined by reputable technicians well used to the vagaries of magnetic tape? If we are going to accuse Raudive of trickery then let's do so openly —and be ready to say how it is done—without invoking explanations which don't explain.

Secondly, the two quotations Mr Fleming gives certainly do not deal a deathblow to the psychic/electronic interaction hypothesis I proposed. Admittedly we have little real knowledge of the laws behind parapsychological activity: but all the evidence says that neither telepathy nor telekinesis are electromagnetic radiation; do not obey the known principles of propagation; and cannot be blocked by any shielding. Therefore the fact that Raudive voices appear on recordings made inside a Faraday cage is quite irrelevant to the present argument.

I will not try to answer the point about the "identical peculiar rhythm" of the recordings because I believe little sense can be said without more facts to go on. But I would point out that brains are not so dissimilar in some of their manifestations as Mr Fleming suggests; for example the alpha rhythms in most normal brains are too alike for it to be possible to distinguish between them by these traces alone; and one of the important things about psychical research is that its evidence suggests all minds are linked "under the surface" in a way we don't under-stand. Therefore it's not too surprising if all brains react with electronic circuitry in much the same way: or at least seem to with our relatively crude equipment. In any case, if Raudive is to be believed. the voices are quite distinctive in pitch and timbre. This is not to say that because the voices are recognised, they may therefore be taken to be those of dead people: I hate to think of any of my dead relatives being forced to utter the kind of imbecilities we have immortalised in "Breakthrough". We do know from research into mediums and their activities that

Practical Electronics November 1972



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AC132         12p         BC157         13p         OC35         45p         2N39260         9p         2N3708           AD140         50p         BC158         13p         OC42         12p         2N29267         9p         2N3709           AD161         33p         BC159         13p         OC42         12p         2N29267         2N3709           AD161         33p         BC159         13p         OC42         12p         2N39264         2N3710           AD162         36p         BD131         75p         OC45         12p         10p         2N3711           AFI14         20p         B5132         75p         OC70         12p         2N3054         Sep         2N4062           AFI16         20p         BF179         32p         OC71         12p         2N3055         60p         ZTX302           AFI16         20p         BF194         15p         OC72         12p         2N3442         ZTX503           BC107         10p         BF195         15p         OC81         12p         2N3702         13p         40362	10p 11p 11p 12p 15p 16p 16p	VEROBOARD         JACK PLUGS AND SOCKETS           21 × 33         21p         16p         Standard screened         18p         25mm insulated         8p           21 × 33         24p         16p         Standard insulated         12p         35mm insulated         8p           21 × 5         24p         24p         Standard insulated         35p         35mm screened         13p           31 × 5         27p         27p         Stereo screened         35p         35mm socket         8p           17 × 24         73p         57p         77p         Stereo screened         13p         25mm socket         8p           17 × 24         73p         73p         73p         74p         D,I,N, PLUGS AND SOCKETS         8p           17 × 35 (plain)         —         82p         2 pin, 3 pin, 5 pin 180°, 5 pin 240°, 6 pin         17           17 × 35 (plain)         —         62p         Plug 12p, Socket 8p.         15m         40°, 6 pin
ZENER DIODES         LINEAR I.C.'s (D.I.L.)         DIL Socket           400mWV 5% 3:3V to 30V, ISp.         709 50p         741, 50p         and I6 pin.           710 50p         748, 50p         748, 50p         and I6 pin.		$\begin{array}{c c c c c c c c c c c c c c c c c c c $
DIODES         SIGNAL           RECTIFIER         SIGNAL           BY127         1250V         IA         12p         OA85           BZV10         800V         6A         25p         OA90           BZY13         200V         6A         20p         OA91           IN4001         50V         IA         7p         OA202           IN4001         1000V         IA         8p         IN4148           IN4007         1000V         IA         12p         BA114	7 5 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 5 5 7 7 5 7 5 7 5 7 7 5 7 7 5 7 7 7 7 7 5 7	Pkt. 50 pins       20p       20p       9V mains power supply. Same size as PP9 battery.         THERMISTORS       THERMISTORS         VAI0555       15p; VAI0665       15p; VAI077       15p; R53       £1-35.         COMPACT CASSETTES—IN PLASTIC LIBRARY BOX       C90       65p       C120       85p.
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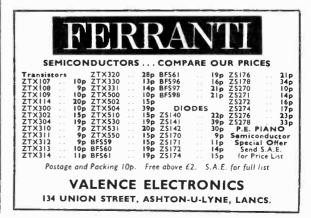
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Practical Electronics November 1972

# A SELECTION FROM OUR POSTBAG

the subconscious mind is capable of generating all kinds of personas (synthetic personalities) to give substance to its longings, hopes, hatreds, etc. That these figments should then escape into the "mental ether" and come to react with electronic equipment I find quite plausible, especially when supported by other experimental evidence.

And there is experimental evidence. If any readers are ready to jib at the notion of mere thoughts directly affecting electronic circuitry I would direct them to the work of Helmut Schmidt, a physicist of international repute. He has devised an apparatus on which trained subjects can both predict and control random events at the sub-atomic level (the decay of Strontium-90 nuclei) at a high level of probability by willing alone. There is a good summary of his work in Arthur Koestler's "The Roots of Coincidence", 1972.

- Peter R. Morton

### Amongst the white noise

Sir—After reading Mr Peter Moreton's letter in the September edition of P.E. on "Psi-Tronics". I wondered if your readers would be interested in similar experiences of my own?

I have a temperamental (!) v.h.f. receiver which I use in a radiotelescope aerial system, and during tuning up of this receiver and its associated preamp to the required band, it was left switched on for long periods producing nothing but what has become to be known as "white noise". The preamp being very difficult to align, I was often switching the signal generator off and on, making adjustments, and trying again. On many occasions, during range testing of my aerial, I am sure I heard the signal generator whilst it was switched off, i.e. all the radio was receiving was the usual v.h.f. spurious noise. This happened so frequently that I had to keep checking that the signal generator was switched off, before I tried to properly "tune in" this figment of my imagination!

I put it to Mr Brown and Mr Morton that the mind has some mechanism whereby it hears amongst the white noise what it wants or *expects* to hear, and that this explains the phenomena, since in my case, the familiar note of my signal generator is unmistakable, and I certainly was fully conscious of the fact I was hearing something that couldn't be there!

I must confess to being still sceptical about such things as the "Raudive Voices", but when it happens to you personally, it's something quite different!

L. M. Newell, Woodbridge, Suffolk.

### **Road choir !**

Sir—In connection with the recent interest shown in "Raudive voices" I should like to give an example of my own subjective interpretation of white noise.

Some years ago, while riding on the cab of a noisy long-distance lorry I was enveloped by the sound of the engine, the sound of other cars on the road and the various vibrations and rattles from parts of the lorry, i.e. a broad band frequency spectrum somewhat resembling white noise. I hummed a pop tune to myself and to my surprise I heard the pop song coming from somewhere inside the cab sung, as it seemed to me, by a choir of outof-tune voices. There was no radio in the cab and it seems that the voices were a purely subjective interpretation of the random noise spectrum reaching my ears. With practice it was possible to hear the voices "sing" any tune at will, there being a period of some seconds between conceiving a tune and actually "hearing" it. Since that time I have met one or two people who have had the same experience. The moral seems to be that if it is possible for some people to hear pop songs in white noise, then it is equally possible for others to hear voices from the dead if they so wish.

An explanation may be as follows. When a memory comes to mind the brain cells that correspond to the sensual experience of that memory become active. If (when the memory is of a noise) the ears perceive white noise simultaneously, then the same mechanism that allows concentration on one conversation when several may be going on will come into effect, only allowing the memory of the noise to come into consciousness. This memory will, however, appear to come directly from the senses since the white noise components corresponding to the noise will be allowed to pass by the brain's "filters".

It may be interesting for psychologists to carry out experiments in this area, and perhaps with other senses. (Is there any capacity to project visual images when in an environment of completely white light for example) perhaps even ending up with "White taste" and "White smell".

I would, of course, be interested to hear other people's interpretations and experiences in this area.

K. Willson, B.Sc., London S.W.16,

### Unconvinced

Sir—I was interested to read in the September issue that the phenomenon of "Raudive Voices" had cropped up again under the general heading of Psitronics. The hypothesis on the origin of this phenomenon suggested by your correspondent, Peter Morton, is most ingenious but is open to criticism on one simple count.

It is now generally recognised that the operation of the brain is related to a measurable form of electrical activity. Thus it is reasonable to assume that telepathic communication is, ipso facto, electrical in origin also. If this assumption is accepted as being substantially correct it becomes increasingly difficult to explain the positive results from experiments into the phenomenon which were conducted with the recording apparatus sealed within a Faraday Cage.

If we accept that the phenomenon exists, and there appears to be plenty of evidence that it does, then it seems that we must resort to the explanation tendered by one of Raudive's original witnesses—"that it transcended the *known* laws of Physics'. All of which leaves a wide open field for future experimenters. Having read "Breakthrough"

Having read "Breakthrough" shortly after its publication a friend and I decided to carry out a number of experiments to prove the existance of the phenomenon for ourselves. My colleague in this, besides being a highly qualified electronics engineer, has had a genuine "feel" for what might be termed supranormal phenomena all his life, and thus felt that everything was on our side.

We built two forms of each type of detector described in the book and combined these, in turn, with six types/models of tape recorder. In addition we tried the microphone and radio methods of recording, but after many hours and several miles of tape had recorded only one voice which could not be directly attributable to spurious "E" propagation or something similar. Our prize voice resisted all attempts of filtration and re-amplification to

# A SELECTION FROM OUR POSTBAG

become intelligible so we decided to retire unconvinced.

I too would be interested to learn of the experiences of others in this field.

#### Douglas Shaw, Eaton Socon, Hunts.

Authoritative opinion appears to be opposed to the view that telepathy is a form of electromagnetic radiation. Dr Grey Walter in his book, "The Living Brain" (1961) states: "if we consider the largest rhythms of the brain as casual radio signals, we can calculate that they would fall below noise level within a few millimetres from the surface of the head".

There is evidently much general interest in this topic. Many readers have written describing their own experiences and experiments, but no one has reported a success in recording "Raudive voices".

### **Retort from Australia**

Sir—l read with great interest the article *Report from Australia* and l am sure that the article was written with the best of intentions in the world. However, the Australians in general tend to gloss over many things, and literally hate comparisons when it comes down to earth, i.e. hard cash! This particularly so in the radio field.

l go green with envy when I read the cost of components, transistors and integrated circuits in the U.K. and I compare them with those available here. For example, take the BC108 listed in P.E. at 10p or 20 cents Australian, the BC108 here is 45 cents and on top of that there is up to 27% sales tax, depending on the State in which you live. This tax is very crippling when you come to build any of the available projects. So my advice to any intending migrant is to do two things: forget most of what the brochures from Australia House tell you, and if you still intend to come, take things as they are and you'll make out. The second thing is to bring out as much gear as you possibly can.

This is a mighty big country, so its a little unfair to classify most radio sets as made for local station reception. There are areas (fewer now than they used to be) of poor reception in the U.K. Well England from top to bottom is approx. 560 miles. Folks here often drive this distance, and more, just to visit friends at a weekend. It has to be a really good quality radio set to pull in inter-state radio stations during the day. At night of course it's not so bad. In the more thickly populated areas this situation may not arise as transmitters are closer together, but it is certainly true of South Australia.

It's all very well to say that Australia is as technically advanced as its contemporaries, but I have found that when dealing with i.c.'s the manufacturers' applications notes are so technical that they are for boffins only—and I include all American manufacturers — so I am indebted to your magazine for *Making The Most Of Logic* and subsequent articles on i.c.'s and most important the way these articles have been presented.

Talking of f.m. 1 bought a German radiogram in Singapore in 1959. I arrived in this country in 1966. F.M. was, as it is now, being talked about but I'll believe it when I see it! I am told they had a few f.m. stations in the Sydney area of N.S.W., but they died through lack of interest as far back as 1966.

Well, this letter was prompted for two reasons, my personal dislike of half truths, and the fact that I'm in the radio field and I did migrate to Australia. So if this does anything to put the record straight I'll be happy.

> P. Hickman, Greenacres, South Australia.

### **Musician's ambition**

Sir—As another of your younger readers I fully agree with Neville Powell's letter in your September issue. It would be nice to see designs for transistorised sound effects apart from the more common tremolo and fuzz, etc. These things are fairly easy for the average constructor to design whereas more complicated effects such as echo, reverb, and waa-waa are harder to design and build.

It is every musician's ambition (1 am no exception!) to own a synthesiser, but the prices of commercially made instruments are out of this world! I am sure it would be possible to make a synthesiser using cheap and easily obtained i.c.'s.

Hoping to see something of this kind in the future issues of PRACTI-CAL ELECTRONICS.

Andrew Copsey, Gildersome, Nr. Leeds.

You will! The P.E. Synthesiser will be on display at the Audio Fair, Olympia, London, October 24-28.

All of the sound effects mentioned in your first paragraph have been covered as constructional projects in past issues of this magazine,

### BACK NUMBERS WANTED

Anyone who can supply the undermentioned are asked to communicate directly with the reader.

#### Any Back Numbers

Mr. E. N. Elia, P.O. Box 297, Baghdad, Iraq.

#### December 1968 Mr. D. W. Jones, Hefindre, Llangefni, Anglesey.

March, April, June 1972 Mr. G. A. Herd, 4, Graig Lwyd, Radyr, Cardiff.

#### October 1971

Mr. T. Svenell, Östra Farmvägen 40B, S-21441 Malmö, Sweden.

#### February 1969

Mr. G. Paine, 21, Estridge Way, Tonbridge, Kent. March 1966 Mr. B. L. Codd, 46, Woodrows, Woodside, Telford, Shropshire.

April 1970 Mr. M. Dowding, 89, Beresford Road, Lowestoft, Suffolk.

February 1971 Mr. B. Whiting, Fellands Gate, Old Leake, Boston, Lincs.

#### November 1964 to March 1965, October to December 1970, February and March 1972

Mr. P. U. Sukhadia, 1–16, Shantinath Bhuvan, 427, Sion Road, Matunga, C.Rly. Bombay-19, India.

We regret that back numbers of Practical Electronics can no longer be supplied. We will try to publish announcements of readers' requirements (without a guaranteed date) free of charge.

#### May 1969 to March 1970

Mr. V. T. Brown, 242, Little Wakering Road, Wakering, Southend SS3 0JN.

March 1968, January 1971, April 1971, June and July 1972 Urs Bachofner, Wulflingerstrasse 92, CH-8400 Winterthur, Switzerland.

#### August 1969

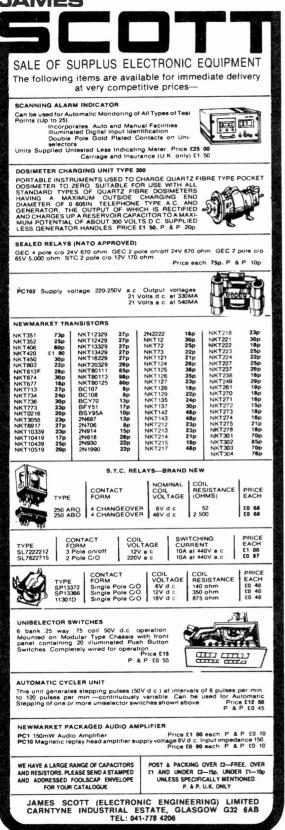
Mr. M. Latter, 45, Wortley Road, W. Croydon, Surrey.

#### April-May 1972

Mr. G. Slack, 87, Upham Road, Swindon, Wiltshire.



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## **Digital I.C. Gates**

A BINARY information signal in logic i.c.s has only two possible states: a logic level 1 of about 4 volts and a logic level 0 of about 0 volts. Each logic circuit produces an output logic level that is a logical function of the input logic levels. These levels are not critical provided that they are clearly recognised as 0 or 1, and small changes in levels cannot accumulate to cause loss of accuracy.

Simple logic functions are achieved by using gates while more complex functions use specifically designed circuits, which may include gates but not necessarily so.

#### TTL SERIES

Several classes of integrated circuit logic are available, including RTL, DTL, TTL and ECL. However, now that TTL logic is available in a variety of speeds and a considerable range of functions, no other type need be considered for general use in this series of articles.

The Series 74 integrated circuits available from many sources are designed and characterised for general purpose high speed digital applications where a high d.c. noise margin and medium power dissipation are important. The series includes gates, bistables, complex logic functions, and many other circuits.

Four compatible TTL families are available to allow switching speed (and hence power dissipation) to be chosen according to the system requirements. The largest and most widely used family is the SN7400 series which has a 10ns propagation delay per gate. (Alternative types are made by various manufacturers and details of equivalents were given in the P.E. I.C. Identichart last month.)

The 74H00 or 74S00 series can be used for minimal propagation delay times, with the 74L00 series reserved tor low speeds. A comparison between these familes is shown in Table 2.1.

The speed differences over the range are not a matter of selection as the circuits differ, although the logic

#### Table 2.1

Series Code	Power Dissipation	Propaga- tion Delay	Approx. Relative Price Factor
SN74L00	1mW per gate	33ns delay	1.8
SN7400			1.0
SN74H00	23mW per gate		2.4
SN74S00	19mW per gate	3ns delay	6.7

functions and levels are compatible. The SN74S00 series, for example, have Schottky barrier-diode clamping to achieve ultra high speeds with the best speed-power product of any high speed logic family.

The Schottky barrier-diode clamping prevents the transistors from going into saturation and eliminates the effect of excess charge storage and subsequent recovery times. These recovery times contribute significantly to the overall propagation delays obtained with conventional saturated TTL circuits.

#### GUIDELINES WHEN USING TTL

*Power Supplies.* Ripple and spikes should be kept to 5 per cent or less and regulation should be maintained to better than 5 per cent. Every five to ten packages should be bypassed to r.f. with ceramic capacitors of 0.1 to  $0.01\mu$ F.

A ground planc is desirable especially in large systems, or make the earth line as wide as possible on a printed circuit board and return both ends of long ground wires to a common point.

*Gates.* Gates should preferably be driven from a low impedance source. If the source impedance is greater than 100 ohms (for example, non TTL circuits) then the input rise and fall times should be less than  $1\mu$ s to avoid instability and oscillation occurring when the gate goes through its active region of operation. Data pulse widths should be 30ns or more.

The unused inputs of AND and NAND gates and unused presets and clears of bistables can be treated as follows:

- (a) They can be tied directly to  $+V_{cc}$  where  $V_{cc}$  is guaranteed to be *always* equal to or less than 5.5V.
- (b) Connected to  $+V_{vc}$  through a resistor of 1 kilohm which protects the input if the supply exceeds 5.5V. Up to 25 unused inputs can be connected to one resistor.
- (c) Connected to an independent supply of between 2.4 and 3.5V.
- (d) Connected in parallel with a used input of the same gate if the maximum fan-out of the driving output will not be exceeded. Each input presents a full load in the logical 1 state to the driving stage.

Completely unused gates can be taken to ground for the lowest power dissipation, or left floating (unconnected).

Unused inputs of NOR gates can be tied to the used input of the same gate (if the maximum fan-out of the driving gate is not required) or returned to ground.

*Bistables.* If a clock pulse is present maintain the preset or clear pulse until the clock goes low. Rise and fall times of the clock pulse should be less than 150ns to aid noise immunity.

In general the input data of a master/slave JK bistable should not be changed while the clock pulse is high, but exceptions are given in manufacturers' data sheets.

#### BASIC GATES

Five logic functions are summarised in Table 2.2. Their interrelationship can be seen by comparing the

**BUFFER** An integrated circuit with a higher fan-out than usual, for driving heavy loads.

**CHARGE STORAGE** Energy stored in a transistor when it is heavily saturated.

**DTL** A logic circuit using diodes for the input coupling to a common emitter amplifier.

**ECL** A logic circuit consisting of an *npn* current mode switch.

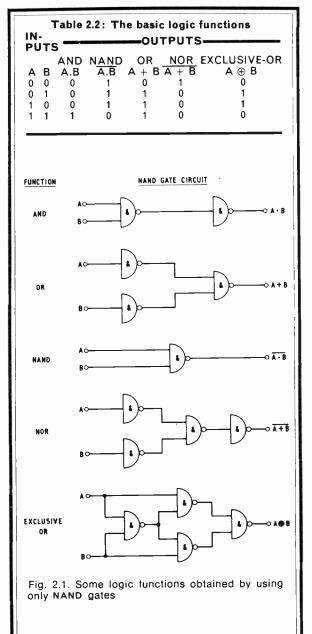
**FAN-OUT** The number of inputs which can be driven by the output of a logic gate.

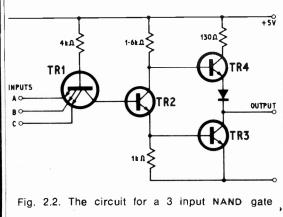
**NOISE IMMUNITY** A measure of the ability of a logic gate to reject noise pulses. It is the smallest 1 level output voltage minus the minimum effective 1 level input voltage, or the minimum effective 1 level input voltage minus the maximum 0 level output, whichever is the smaller.

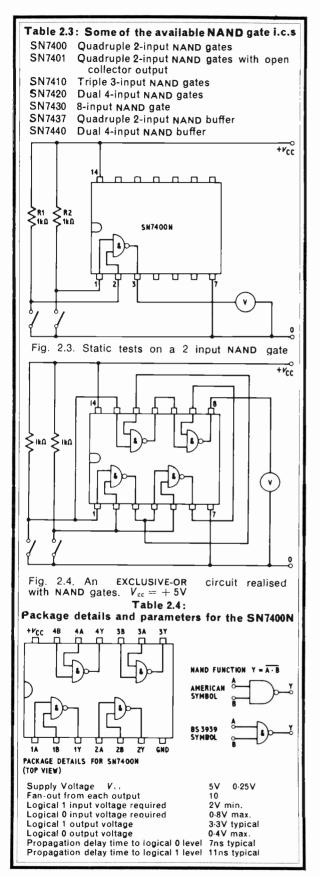
**PROPAGATION DELAY** A measure of the time taken for a change in logic level to be transmitted through an element.

**RTL** A logic circuit using resistors for the input coupling to a common emitter amplifier.

**TTL** A logic circuit having all inputs connected to the multiple emitters of a single, common base connected transistor.







It can be seen that the AND function gives a 1 when both A and B are at 1, and so on. The EXCLUSIVE-OR function has a symmetry of its own and is sometimes considered as a separate basic logic function.

Most of the logic functions can be achieved by using only NAND or only NOR gates at the expense of speed and package count, so that for experimental work it is not necessary to maintain a complete stock of all types. This is shown in Fig. 2.1.

#### NAND GATES

The internal circuit for a three-input TTL NAND gate is shown in Fig. 2.2. TR1 is a multiple emitter transistor. If one or more emitters are grounded the transistor is forward biased, the collector is at a low potential, and TR2 is turned off. This allows TR3 to turn off and TR4 to conduct, resulting in a logic 1 output level.

If, however, all inputs are high the base-collector junction of TR1 will conduct, forward biasing TR2. This turns TR4 off and TR3 on, giving a logic 0 at the output.

The standard TTL gate has a 10ns propagation delay time, a fan-out of about 10 and a noise immunity of at least 1V, making it suitable for most applications. Table 2.3 lists some of the common NAND gates available.

The SN7400 is a quadruple two-input positive NAND gate, package details and parameters are shown in Table 2.4. Although the logic functions can be worked out from the circuit in this case, for more complex packages the corresponding truth table would have to be used.

### EXPERIMENT TO VERIFY THE NAND FUNCTION

For experimental work it is convenient to wire several dual-in-line type i.c. holders to pins on a patch board or experimental wiring board; six holders would be enough for many applications and facilitates replacing i.c. packages for other purposes.

The circuit in Fig. 2.3 could be made up to confirm the truth table, the meter indicating go or no-go conditions (or 1 and 0).

### EXPERIMENT TO VERIFY THE EXCLUSIVE-OR FUNCTION

The circuit is shown in Fig. 2.4. This function gives an output when A and B are not equal. If one input is inverted the overall function will give an output when A and B are equal. Both arrangements are similar to the upstairs-downstairs light switching found in many homes where a lamp can be switched on or off from either of two places.

Although in this case an EXCLUSIVE-OR package is available (the SN7486) the circuit shows how a simple logic function can be obtained using NAND gates when a suitable package is not to hand.

This article has concluded with two experiments illustrating the static characteristics of the NAND gates. Future articles will deal with further applications of the TTL logic families.

#### Next month: Basic operational amplifiers

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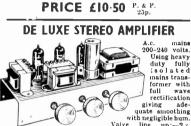
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200/240 V A.C. input. Four awitched fully smoothed D.C. outputs giving 6V and 74V and 9V and 12V at 1 amp continuous (1) amp intermitten). Fitted insulated output terminals and pilot iamp indicator. Hammer finish metal case, overall size 6\*  $\cdot$  34\*  $\cdot$  24\*. Suitable for Transistor Radios, Tape Recorders, Amplifiers, etc., etc. Ready built and tested.

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F. & F. 30p.
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Pri. 200/240V. Sec. 9-0-9 at 500nA. 70p. P. & P. 13p.
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GENERAL PURPOSE HIGH STABILITY TRAN-SISTOR PRE-AMPLIFIER. For P.U. Tape, Mike, Guitar, etc., and suitable for use with valve or transistor equipment. 9-18V. Battery or from H.T. line 200/300V. Frequency response 15H-25KHZ. Gain 26dB. Solid encapsulation size  $l \chi < l j < l in.$ Brand new — complete with instructions. Price880. P. & P. 130.Brand new - com 889, P. & P. 13p.

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QUALITY RECORD PLAYER AMPLIFIER MK II QUALITY RECORD FLATER AMPLIFIEM MR II A top-quality record player amplifier employing heavy duty double wound mains transformer, ECC83, ELS4, and rectifier. Separate Baes, Treble and Volume controls. Complete with output transformer matched for 3 ohm speaker. Sizer Jin. w. ~ 31. < 6 h. Ready built and teated. PRICE 43-75. P. & P. 40p. ALSO AVAILABLE mounted on board with output transformer and speaker ready to hit cabinet below. PRICE 44-85. P. & P. 50p. DE LUXE QUALITY PORTABLE R/P CABINET MK H De LOAR GUALIT FORTAGLE R/F UNDERST MAI Uncut motor board size 144 - 12in. clearance 2in. below. §4jin. above. Will take above amplifier and any B.S.R. or GARRARD changer or Single Piayer (except AT60 and SP25). Size 18×15×8in. PRICE 24.75. P. & P. 50p.

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Beautifully made teak finish enclosure with most attractive Tygan-Vynair front. Size 1641n blgb  $\times 1041n$  wide  $\times 51$  deep. Fitted with E.M.I. Ceramic Magnet 13in  $\times 81n$  base unit, two R.F. tweeter units and crossover. Power handling 10W. Available 3, 8 or 15 ohm impedance.

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Adjustable headband with comfortable flexifoam ear-muffs. Wired and fitted with standard stereo jin jack plug. Frequency response 30-15,000Hz. Matching Automatic Wired and fitted with statutate wired and fitted with statutate and fitted with statutate and plug. Frequency response 30-15,000Hz. Matching impedance 8-16 ohms. Easily converted for mono. PRICE

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A really first-class Hi-Fi Stereo Amplifier Kit. Uses 14 transistors including Silicon Transistors in the first five stages on each channel resulting in even lower noise level with improved sensitivity. Integrated pre-amp with Bass, Treble and two Volume Controls. Suitable for use with Ceramic or Crystal cartridges. (Very simple to modify to suit magnetic cartridge-instructions includel). Output stage for any speakers from 6 to 15 ohms. Compact design, all parts supplied including drilled metal work, high quality ready drilled fibre glass printed circuit board, smart brushed anodised aluminium front parel with matching knobs, wire, solder, nuts, bolts-no extras to buy. Simple step by step, instructions enable any constructor to build an amplifer to be proud of. Brief specification: Power output 14W r.m.s. per channel into 6 homs. Frequency response ± 3dB 12-30,000Hz. Basativity better than 80mV into 1M Q approx. to +12dB. Treble cut approx. to -10dB. Megalik techback 10dB over main any. Nowr require-ments of the start of the specification approx. The -10dB. Treble designed for any per senterment on any and and mark life the public for any constructor to may and mark ints (der public designed for any constructor to may and sets ints of the specification approx. The sentermine and the fitted for the form the specification of the sentermine and the set of the form the specification approx. The sentermine approximate and the fitted form the specification of the sentermine and and the set of the set of the set of the sentermine the sentermine and the set of the sentermine and the set of the sentermine t 21" high.

24<sup>\*</sup> high.
 Paily detailed 7-page construction manual and parts list free with kit or send 18p plus large 8.A.E.
 PRICES AMPLIFIER KIT, \$1050 P.& P.15p. (Magnetic input component 30p extra) POWER PAYK KIT, \$3 P.& P.30p. CABINET, \$3 P.& A 9.30p.
 (Post Free if all units purchased at same time). Full after sales service. Also available ready built and tested, \$2050. Post Free.

22050, Poat Free. Note: The above amplifier is suitable for feeding two mono sources into inputs (e.g. mike, radio, twin record decks, etc.) and will then provide mixing and fading facilities for medium powered Hi-Fi Discotheque use, etc.



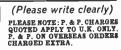
3-VALVE AUDIO AMPLIFIER HA34 MK II ANPLIFIER HASA MK II Designed for Hi-Fi reproduc-tion of records. A.C. Mains optated heavy gauge metal chasis, size 74in w. 44in. 4. Honerportate SCC83, EL84, E280 valves. Heavy duty, double wound mains transformer and output trans-former ranched for 3 ohm speaker. Separate volume control and now with improved wide range tone controls giving bass and treble lift and cut. Negative feelback line. Output 44 watts. Front panel can be detached and leads extended for remote mounting of controls. Complete with knobs, valves, etc., wired and tested for only 2475. P. & P. 35p. HEL "FOUR" AMPLIFIER EKTS, Similar in appearance to

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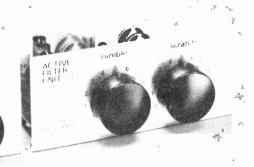






# Sinclair Project 60





# Active Filter Unit

The value of an efficient filtering system cannot be over emphasized in these days of very high quality reproduction since there are so often occasions where its use can mean the difference between comfortable and uncomfortable listening. On the low pass side the Sinclair A.F.U. will effectively reduce hiss from radio or tape, cut out heterodyne whistles on A.M. reception, greatly reduce record surface noise and other imperfections; on the high-pass side it will cut out motor rumble and other spurious low frequency intrusion. The unit is for use between pre-amp (including tape pre-amps) and power amplifiers, and operates in two sections, both stereo. The cut-off frequencies are continuously variable, and since attenuation in the rejection band is rapid (12dB/octave) there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is as easy to mount as the stereo 60 pre-amp/control unit which it matches in styling, along with the Stereo FM Tuner.

Built and tested post free £5.98

#### SPECIFICATIONS

The A.F.U. employs two Sallen and Key type active filter stages, one rumble (high pass) and one scratch (low pass). The two stages use complementary transistors to minimise distortion

Supply voltage: 15 to 35 volts. Current 3mA maximum

Gain at 1kHz: Filters flat 0.98 (-0.2 dB) HF cut off: (-3dB) variable from 28 kHz to 5 kHz at 12 dB/octave.

LF cut off: (-3dB) variable from 25Hz to 100Hz at 12dB/octave.

Distortion: at 1kHz (35 volt supply) 0.02% at rated output.

### Super IC.12 Integrated circuit high fidelity amplifier



Having introduced Integrated Circuits to hi-fi constructors with the IC.10, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super IC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up, F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC, make possible. It is the equivalent of a 22 tran-



sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC.12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Complete with free manual and printed circuit board.

#### SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak). 6-80. Frequency Response: 5Hz to 100KH2±1d8. Total Harmonic Distortion: Less than 1% (Typical 0-1%) at all output powers and frequencies in the audio band (28V). Load Impedance: 3 to 15 ohms. Input Impedance: 250 Kohms nominal. Power Galn: 90dB (1,000.000,000 times) after feedback. Supply Voltage: 6 to 28V. Quiescent current: 8mA at 28V. Size: 22×45×28mm including pins and heat sink.

Manual available separately 15p post free,

With FREE printed circuit board and 40 page manual. **£2.98** Post free



the Project 605 method of connecting. Complete Project 605 pack with comprehensive manual, post free Everything you need to assemble a superb 30 watt high fidelity stereo amplifier without having to solder.

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## the world's most advanced high fidelity modules

PZ.5 30 volts unstabilised £4.98

£7.98

£7.98

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PZ.6 35 volts stabilised

PZ.8 45 volts stabilised less mains transformer)

PZ.8 mains transformer

#### Z.30 & Z.50 power amplifiers

Built, tested and guaranteed with circuits and instructions manual z.30 £4.48 z.50 £5.48

The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (80) and all lower outputs. Whether you use Z 30 or Z 50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that Z.50s and Z.30 may be used in a far wider range of applications.

SPECIFICATIONS (2.50 units are interchangeable with 2.30s in all applications) .-- Power Outputs: Z.30 15 watts R.M.S. into 8 ohms using 35 volts. 20 watts R.M.S. into 3 ohms using 30 volts.
Z.50 40 watts R.M.S. into 3 ohms using 40 volts. 30 watts R.M.S. into 3 ohms using 50 volts.
Frequency response: 30 to 300,000Hz ± 1dB, Distortion: 0.02% into 8 ohms. Signal to noise ratio: better

than 70dB unweighted. Input sensitivity: 250mV into 100 Kohms (for 15w into 8Ω). For speakers from 3 to 15 ohms impedance Size: 14 x 80 x 57mm

### Stereo 60 Pre-amp/control unit

Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount.

SPECIFICATIONS—Input sensitivities: Radio – up to 3mV. Mag. pu. 3mV· correct to R.I.A.A. curve ±1dB 20 to 25.000 Hz Ceramic pu – up to 3mV Aux – up to 3mV. Output: 250mV. Signal to noise ratio: better than 7ddB. Channel matching: within 1dB. Tone controls: TREBLE+12 to --12dB at 10KHz: BASS 

### Project 60 Stereo F.M. Tuner

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

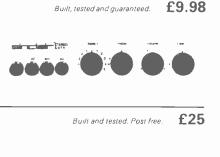
SPECIFICATIONS--Number of transistors: 16 plus 20 m I.C. Tuning range: 87 5 to 108MHz. Sensitivity: 7μV for lock-in over full deviation Squelch level: Typically 20μV Signal to noise ratio: >65dB. Audio frequency response: 10Hz - 15KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2μV. Cross talk: 404B. Output voltage: 2 x 150mV R M S maximum Operating voltage: 25-30VDC Indicators: Stereo on, tuning Size: 93 x 40 x 207mm.

### Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 or PZ.8 where a stabilised supply is essential

Typical Project 60 applications

System	The Units to use	together with	Units cost
Simple battery record player	Z.30	Crystal P.U., 12V battery volume control, etc.	£4.48
Mains powered record player	Z.30, PZ.5	Crystal or ceramic P U. volume control, etc.	£9.45
12W. RMS continuous sine wave stereo amp for average needs	2 x Z.30s, Stereo 60; PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc	£23.90
25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers	2 x Z.30s, Stereo 60; PZ.6	High quality ceramic or magnetic P.U., F.M, Tuner, Tape Deck, etc.	£26.90
80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms)	2 x Z.50s, Stereo 60 ; PZ.8, mains transformer	As above	£34.88
Indoor P.A.	Z.50, PZ.8, mains transformer	Mic , guitar, speakers, etc , controls	£19.43
F.M. Stereo Tuner (£25) & A	.F.U. (£5.98) may be	added as required.	







#### Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionics Ltd., you are dissatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantee in co-operation with Sinclair Radionics Ltd.

Each Project 60 module is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise in normal use, we will service it at once and without any charge to you, if it is returned within two years from the date of purchase. Outside this period of guarantee a small charge (typically £1.00) will be made. No charge is made for postage by surface mail. Air Mail is charged at cost

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First time marketed in England, this exciting electronic instrument the "Decoytronic" is the one responsible for starting the electronic fish-lure rage in the U.S.A. just recently. Uses ingenious double-action method of attracting all kinds of fish from hundreds and hundreds of feet away . . . saltwater or freshwater. Why this device is so fantastically successful is because it actually initiates the sound of net insects milling about the surface. These sound waves spread out hundreds of feet in all directions. Although this peculiar sonic frequency sound like much to you-to all the fish in the area it's their dinner bell! But that's not all . . . electroluminescence neur tubes cunningly flicker intermittently, penetrating the area for hundreds of feet around, at a frequency fish are unable to resist. The fish mistake this flickering glimmer for the soft phosphorescence glow given off by Plankton . . . a favourite delicacy of most fish! All you do is switch on, lower into the water (it's completely water resistant) allow around 5 minutes-then start reeling 'em in. We'll bet you won't be able to reel em in fast enough. Self-contained batteries last ages-cost pence. Kit of all parts including special easy as a.b.c. directions only \$2.50+ 25p P. & P. (Parts available separately.)

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eimpily blow air over mouth-organ tupe rerds, etc. Fully tran-sistorised. SELF-CONTAINED LOUD-SPEAKER. Fiffeen separati. keys span two full octares-play the 'Fellow Rose of Texas', play 'Silent Night', play 'Luid Lang Syne'', cle. etc. You have the thrill and excitement of building it together with the pleasure of playing a real. live, portable electronic organ. NO PREVIOUS KNOWLEDGE OF ELECTRONICS MEDED. No soldering necessary, simple as ABC'to make. Anyone short creating following the fully illustrated tere-by-step. simple instructions. OKLY \$3 25 + 25p p & p for kH. Including case, standard battery (parts available separately). Have all the pleasure of making it yourself. finish with an exciting gift for someone.

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nights? Do you wake up in the night and can't get off to sleep again? Would you like to be gently soothed off to satisfying sleep every night? Then build this ingenious electronic sleep inducer. It even stops by itself so you on't have to worry about it being on all night! The loudspeaker produces soothing audio-frequency souths, continuously repeated—but as time goes on the sound gradually become less and less—unit they eventually cease altogether, the effect it has on people is anazingly very shuilar to hypnosis. A control is provided for adjust-ing the length of times, etc., all transistor.

hypnosis. A control is provided for adjust-ing the length of times, etc., all transistor, can be built by anyone over 12 years of age in about two hours. No knowledge of electronics or radio needed. Extremely simple, easy-to-follow, step-by-step, fully illustrated instructions included. No solder-ing necessary. Works of standard batterles, extremely economical. Size only 3in x 41m x 12m-take it anywhere. Kit includes case, muts, wire, screws, etc. SEND 83-25 + 25p P. & F. (Parts available

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Radio Conree. No ex-perience needed. Parta including simple instructions for each design. Illustrated step-by-step plans, all transistors, loudspeaker, personal phone, knobs, screws, etc., all you need. Preventation box 45p extra as illus, (if required) (parta available separately) no soldering necessary. Send 18:45 + 20p p & p.

#### SOOTHE YOUR NERVES **RELAX WITH THIS AMAZING** RELAXATRON

CUTS OUT NOISE POLLUTION-SOOTHES YOUR NERVES! Don' underestimate the uses of this fantastic new design—the RELAXATRON is RELAXATRON is hasically a pink noise generator. Besides being able to mask out

being an be directly traced to the sound of being can be directly traced to the sound of falling raindropal -as well known type of pink noise. IF YOU WORK IN NOISY OR DISTRACTING SURROUNDINGS, IF YOU HAVE TROUBLE CON-CENTRATING, IF YOU FEEL TENSED, UNABLE TO RELAX—then build this fantastic Relaxatron. Once used you will never want to be without it—TAKE IT ANYWHERE, Uses standard PP3 batteries (current used so small that battery life is almost shelf-life). CAN BE EASILY BUILT BY ANYONE OVER 12 YEARS OF AGE using our mique, step-by-step, fully illustrated plane. No soldering necessary All parts including case, a pair of crystal phones. Components, mits, screws, wire, etc., no soldering. 



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x 1311----take case, nuts, 83-25 + 25p separately.)

ONLY £2.75 Everyone's heard the weird, wonderful, but

CONLY 42:75 Everyone's heard the weird, wonderful, but beautiful music used in Science Piction with the standard PB human set of the standard provided by a little known electronic device measuring only a few inches ... called a three min. All you have to do is switch on your radio three and the standard provided by a little practice by the standard provided by a standard provide base standard provided by a standard provide base st available separately.)

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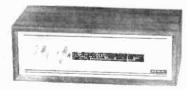
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BN7406 0-80 0	-24 SN7413 0-	80 0-27	CA3045	1.42 1.17	CA3093E	
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VID	EST RA	NGE	5	SEM		NDU	сто	RS 8	& C	OMPONENTS	BRAND NEV
301	- 15p <sub>1</sub> 2N2923	12p+1	TR/ 2N4930	4NSIS		<b>24</b> p <sub>1</sub> BC30	34n 1	BF255	15p	D.C. PANE	
302 303	15p 2N 2924 25p 2N 2925		2N 4931 2N 5172	£2-25 A £2-70 A 8p A	F117	23p   BC30 20p   BC30	2 27p	BF257 BF258		MR38 SER1ES-FACE SIZE 42 × 421 50μA, <b>£2·00</b> ; 100μA, <b>£1·874</b> ; 200μ	nn. ALL PRICES FOR $1-9$ PIECE A, $\$1.75$ ; 500 $\mu$ A, $\$1.50$ ; 50-0-50 $\mu$
306 309	80p 2N2926 80p , Green	100	2N 5174 2N 5175	22p A 26p A	F118 F121	50p BC30 22p BC30	43p	BF259 BF261	48p 23p	50μΑ, £2.00; 100μΑ, £1.87]; 200μ. £1.87]; 100-0-100μΑ, £1.75; 500-0- £1.60; 10mΑ, £1.60; 50mA, £1.60 [Α, £1.60; 5Α, £1.60; 10V, £1.60	500μΑ, <b>±1·60</b> ; ImA, <b>±1·60</b> ; 5m <b>0</b> ; 100mA, <b>±1·60</b> ; 500mA, <b>±1·6</b>
44A 45B	25p Yellow 25p Orange	10pi:	2N 5176 2N 5245	32p A 48p A	F124	22p BC30 20p BC30	7A 9p	BF264 BF270	£1.45 25p	11A, £1.60; 5A, £1.60; 10V, £1.60 £1.60; 500V, £1.60.	; 20V, \$1.60; 50V, \$1.60; 300
71 74	15p 2N 3053 15p 2N 3054	15p ;	2N5190 2N5191	92p A 96p A	F126	19p BC30 20p BC30	5 8p	BF271 BF272	21p 53p		
81 17	22p 2N 3055 20p 2N 3390	50p 2 20p 2	2N5192 2N5193	£1.24 A £1.01 A	F139	80p BC30 25p BC30	3B 8p	BF273 BF274	25p 28p	SLIDER POTE	ND LIN.
09 74	49p 2N 3391 21-40 2N 3391 A	20p	2N 5194 2N 5195	£1.10 A £1.46 A	F172	25p BC30 42p BC30	A 9p	BF457 BF458	46p	SINGLE GANGED 40p	DOUBLE GANGED 6
76 74	75p 2N 3392 75p 2N 3393	13p 2	2N 5245 2N 5457	43p A 30p A	F179	45p BC31 50p BC32	30p	BF459 BF821	57p £2.10		
35 51	75p 2N3394 95p 2N3402	120	2N5458 2N5459	33p A 33p A	F186	40p BC32 35p BC33	s 19p	BFS21A BFS28	£2-30 92p	DIODES &	RECTIFIERS
76 84	<pre>£1 2N3403 98p 2N3404</pre>	19p 3	3N 128 3N 138	63p A £1.37 A	F239	36p BC33 63p BCY3	17p	BFS61 BFS98	27p 28p	1N914 7p 18940 5p 1N916 7p AA119 10p	BAX13 124p BYZ13 2 BAX16 124p OA5 17
88A 04	40p 2N3405 20p 2N3414	27 p 3	3N 139 3N 140	£1.36 A 76p A	F279	47p BCY3 47p BCY3	1 30p	BFW10 BFW11	61p 61p	1N 4007 20p AA129 10p IN 4148 7p AAZ13 10p	BAY18 174p OA9 1 BAY31 74p OA10 22
56 56A	75p 2N3415 75p 2N3416	10p 3	3N 141 3N 142	69p A 54p A	FY42	65p BCY3 55p BCY3	3 25p	BFW15 BFX13	75p 23p	IN5145 22-37 AAZ15 121p 1844 9p AAZ17 121p	BAY38 25p 0A47 BY100 174p 0A70
57A 91	80p 2N3417 £8-25 2N3570	21p £1.25	3N143	64p A 79p A	L102	75p BCY3 70p BCY3	8 40p	BFX29 BFX30	25p 25p	IS113 15p BA100 15p IS120 15p BA102 22ip	BY103 224m OA73 1
84 91	26p 2N 3571 34p 2N 3572	£1.12	3N 153 3N 154	74p A 74p A	SY26	25p BCY4 30p BCY4	0 <b>50p</b>	BFX37 BFX44	80p 83p	18121 14p BA110 324p 18130 8p BA115 74p	BY122 374p OA79 BY126 15p OA81 BY127 174p OA85 1
96 97	15p 2N3702 15p 2N3703	10p 3	3N159 3N187		SY28	25p   BCY4 27p   BCY5	3 15p	BFX63	£2-48	IS131 10p BA141 32 p IS132 12p BA142 32 p	BY164 52 p 0A90 BYX10 22 p 0A91
98 99	25p 2N3704 29p 2N3705	10p 3	3N 200 3N 201	£2.07 A £1.05 A	SY50	20p BCY5 35p BCY6	9 <b>19</b> p	BFX68 BFX84	80p 24p	18920 7p BA144 124p 18922 8p BA145 20p	BYZ16 35p 0A95 BYZ11 321p 0A200 1
06 06A	10p 2N3706 12p 2N3707	90p 10p	514201	A	SZ21	51p BCY6 1-25 BCY7	7 82p	BFX85 BFX86 DBX87	29p 24p	18923 12p BA154 124p	BYZ12 30p T1V307 5
08	13p 2N3708 38p 2N3709	70p		B	SC107	10p BCY7 10p BCY7	1 20p	BFX87 BFX88 BFX89	25 p 20 p	MAINS TRANSFORMERS "VARNIS	
11 18	80p 2N 3710 21p 2N 3711	90p 4 90p 4		50p H	C109	10p BCY8 10p BCY8	7 £3-47	BFY10	45p 85p	1 amp Charger. Sec. 0-3:5-9-18V (P 2 amp Charger. Sec. 0-3:5-9-18V (P	& P. 224p.) 21-4
18A 20	30p 2N3712 50p 2N3713	95p 4 £1.08 4	40309	26p B	C114	12p BCY8 15p BCZ1	9 <b>80p</b>	BFY11 BFY17	45p 90p	1 amp (Douglas) MT103 Sec. tapping 2 amp (Douglas) MT104 Sec. tapping	s from 6V to 50V <b>\$2-1</b> s from 6V to 50V <b>\$3-1</b>
21	500 2N3714 150 2N3715	£1.15 4 £1.23 4	40360	35p B	C116	15p BCZ1 18p BD11	i 40p	BFY18 BFY19 BFY20		Post and packing 25p. 6 amp (Douglas) MT107 Sec. tapping	from 6V to 50V
6	17p 2N3716 80p 2N3773	£1.30 £2.35	40362	40p B	C117	20p   BD11	6 7 <b>5</b> p	BFY29	50p 40p	Post and packing 37 <sup>1</sup> 2p. Various other Transformers ranging	from #A to 5A in stock.
19 29	20p 2N 3774 14p 2N 3775	£3.33 4 £4.19 4	40389	37p B	C119	27p BD12	3 80 p	BFY37 BFY41	20p 43p	TRIACS	BRIDGE RECTIFIERS
30 )90	14p 2N3776 28p 2N3777	£5.95 4 £4.84 4	40395		C123	25p   BD13	0 50 pl	BFY43 BFY50	62p 16p	SC41A         85p         SC35D           SC41B         85p         SC36D	\$1 A. PIV A. PIV 90p 1 100 37p 4 50 4
)91 131	24p 2N 3778	£2.25 4	40407	81p   B	C126	15p BD13 20p BD13	2 75p	BFY51 BFY52	16p 16p	SC50D . \$1.90 SC45D 4	1.25         1.4         140         57p         4         100         7           1.50         2         50         32p         4         400         8
32 184	20p 2N3779 20p 2N3780 £1-27 2N3781	£3.15 4 £4.50 4 £8.67 4	40409	49p B	C134	30p BD13 11p BD13	6 44p	BFY53 BFY56	15p 84p	40486 85p 40429	80p 2 200 41p 6 200 8
102 103	16p 2N 3782 16p 2N 3789		40411		C136	11p BD13 15p BD13	8 <b>55p</b>	BFY64 BFY75	41p 40p		1 37 2 400 46p 6 400 £1-
304 305	20p 2N3790 20p 2N3791	£1.90 4	40468A	57p B 35p B	C138	15p BD13 20p BD14	0 78p	BFY76 BFY77	22p 24p	Economy Range Triacs TC4/10 (Pressfit) 4 amp 100 PIV	D1ACS 65p MPT20 4
06 107	22p 2N 3792 22p 2N 3794	£2.20 4	40600 40601	57p B 55p B	C141	30p BDY1 84p BDY1	1 £1.50	BFY78 BFY90	36p 55p	TC4/40 (Pressfit) 4 amp 400 PIV	75p MPT28 87 874p MPT32 87
08 09	25p 2N 3819 25p 2N 3820		40603	40p B 49p B	C143	18p BDY1 21p BDY1	8 £1.75	SRY39 BSX19	30p 13p	ST2 DIAC	21p MPT36 87
83 07	90p 2N3823 24p 2N3824		10604 10636 10673	45p B 33p B 60p B	C145	24p BDY1 21p BDY2 10p BDY3	0 92p	BSX20 BSX21	14p 20p	P.E. SCORPIO IG	NITION SYSTEM
613 531	20p 2N 3826 35p 2N 3854	28p A	AC107 AC113	30p B 16p B	C148	9p BDYe 10p BDYe	0 900	B8X26 B8X27	84p 34p	COMPLETE KIT £1	0.00 P. & P. 50p
637 538	80p 2N 3854A 87p 2N 3855	16p A		16p B 20p B	C153	10p BDYe 11p BF11	2 75p	BSX 28 BSX 29	25p 47p		
6711 701	£1 2N 3855A £1.10 2N 3856	16p A 16p A	AC121 AC126	11p B 20p B	C157	9p BD11 10p BF11	7 <b>43</b> p	BSX 30 BSX 59	68p 78p	SUB-MIN EL	ECTROLYTIC 6p eac
702 711	£2.15 2N 3856A 17p 2N 3858	16p A	AC127 AC128	20p B 20p B	C159	10p   BF12 11p   BF12	25p	BSX60 BSX61	54p	Values $(\mu F/V)$ ; 0.64/64; 1/40; 1.6 6.4/6.4; 6.4/25; 8/40; 10/16; 10/64; 1	25; 2.5/16; 2.2/63; 4/10; 4/4
393 102	84p 2N3858A 80p 2N3859	16p A 16p A	AC141K AC142K	30p B 25p B	C167B	11p BF12 10p BF12	25p	BSX76 BSX77	15p	25/25; 32/10; 32/40; 32/64; 40/16; 5 80/16; 80/25; 100/6 4; 125/10; 200/6 4	0/6-4; 50/25; 50/40; 64/10; 80/2-
147 148	70p 2N3859A 60p 2N3860	16p A	AC151V AC152V	12p B 15p B	C168C	10p BF15: 11p BF15:	200	B8X78 B8W70	25p 28p		
192 192 a	40p 2N3866 40p 2N3877	25p A	AC153 AC153K	22p B	C169C C170	11p   BF154	16p	BSY24 BSY25	20p 15p	THYRISTORS PIV 50 100 200 300 400	MULLARD C280 M/FOIL CAPACITOR
93 93A	40p 2N 3877A 42p 2N 3878	26p A £1.22 A	AC176	16p B	C172 :	13p BF159	27p	BSY26 BSY27	17p	1A 25p 27ip 37ip 40p 47ip 4A 40p 45p 55p - 80p	0.01, 0.022, 0.033, 0.047, 3p; 0.06
94 94A	27p 2N 3879 30p 2N 3900	£1-91 A 20p A	AC176K AC187K	17p B 17p B		10p BF161	85p	BSY28 BSY38	15p 15p	7A — 871p 921p — £1-121 TIC47 0-6A 200 PIV 58p	0·1, <b>4p</b> each. 0·15, 0·22, 0·33, <b>5p</b> each.
95 95 A	87p 2N 3900A 18p 2N 3901	32p A	AC188K ACY17	28p B 25p B	C183 C183L	9p BF160 9p BF167	35p	BSY39 BSY31	15p	Also 12A 100 P1V, 75p; 2N3525 at 85p.; 2N4444 £1-91	0-47 9p 0-68 11p
218A 219	18p 2N 3903 20p 2N 3904	17p A	ACY 18 ACY 19	15p B 20p B	C184L j	110 BF173	19p	BSY52 BSY53	25p 25p	VEROBOARD 0-15 0-1	1μF 14p 1·5μF 21p
19A 20	20p 2N 3905 20p 2N 3906	21p A 20p A			C186 🛔	25p   BF178 25p   BF179	25 p	BSY54 BSY56	80p 79p	Matrix Matrix	2·2μF <b>25</b> p
21 21 A	20p 2N 4036 22p 2N 4037	40p A 85p A	ACY22	10p B	C204 1	11p BF180	35p ]	BSY65 BSY78	15p	21 in × 5in 25p 25p	WIRE-WOUND RESISTORS 2.5 wait 5% (up to 270 ohi
22 22 A	20p 2N 4058 25p 2N 4059	12p A 90p A	ACY39	42p B 47p B	C206 1	11p BF18: 10p BF18:	30p ]]	BSY79 BSY790	40p	3‡in × 3‡in 25p 25p 3‡in × 5in 30p 29p 5in × 17in (plain) 83p —	only), 7p. 5 watt 5% (up to 8.2kΩ only), 1
68 69	11p 2N4060 12p 2N4061	11p A 11p A	ACY40	17p B	C208	9p BF184	17p	BSY95A BU104	9p	Vero Pins (bag of 36), 20p.	10 watt 5% (up to 25kΩ only 10p,
69A 46	15p 2N 4062 41p 2N 4302	11p A	CY44 D136V	25p B 84p B	0211 8	0p BF194 0p BF193	14p	BU105 2111	20 OF	Vero cutter, 45p; Pin insertion Tools (0.1 and 0.15 matrix) at 55p.	POTENTIOMETERS
11	2N4303 12p 2N4913	82p A 80p A	D140 D142	49p B 50p B	C212L 1 C214L 1	2p BF196 4p BF197	15p (	0407 0424	18p 15p	OPTOELECTRONICS	Carbon:
12 13	12p 2N4914 17p 2N4915	87p A 95p A	D143 D149V	45p B 50p B	C236 ] C237	6p BF198 8p BF199	15p (	C425 C426	36p 25p	MINITRON 3015F 7-SEGMENT INDICATOR (SEE P.E. MARCH	Log. or Lin., less switch, 16p. Log. or Lin., with switch, 25p. Wire wound Pots (2W) 29p.
14 04	17p 2N4916 18p 2N4917	11p A 17p A	D149V D150	£1-23 B 55p B	C238 C239	8p BF200 8p BF224	35 p	0428 0444	25p	INDICATOR (SEE F.E. MARCH 1972) 22 GNP-7AH COLD CATHODE	Wire-wound Pots (3W), 38p. Twin (Janged Stereo Pots, Lo and Lin 40n
04A	25p 2N 4919	50p A 58p A	D161 D162	33p B 35p B	0251 \$	80 BF225	J 19p	D40N1 D40N3	55p	SIDE-VIEW NIXIE TUBES 75p. TIL-209 RED LIGHT EMITT-	and Lin., 40p.
05 05 A	25p 2N4920 28p 2N4921	60p A 50p A	$D161 \\ D162 $	80p B	C253 \$ C257	8p BF237 8p BF238	22p	GETIII	45p 20p	ING DIODE (TEXAS INST.) 85p	PRESETS (CARBON) 0-1 Watt 6p
06 06A	18p 2N4922 2N4923	55p A 60p ,	DZ11 DZ12	£1 50 B	C258 C259 1	8p BF244 1p BF245	16p	GETI14 GET115	20p 50p	RESISTORS	0-2 Watt 6p 0-3 Watt 74p
	18p 2N4926 18p 2N4927	90p A	F106	24p B	C261 8	BF246	48p	GET119 GET120	85p 25p	Carbon Film 1 watt 5%, 1p. 1W & W are	SLIDER POTENTIOMETERS
07	25p 2N4928	£1.80 A £2.23 A		35p B	C263 🖇	8p BF247 4p BF254	49p (	GET535 GET536	20p 20p	1 watt 5%, 1p. E24 series. 4 watt 5%, 14p. 4W, 1W, 4.2W	SINGLE GANGED ONLY LIN: 1k, 2.5k, 20k, 25k, 100
07 07A 13		90.00 J V	C. T. T. H	200.00		AB 1 DT 201					
07A 13	75p 2N 4929 Matcl king 13p per orde	ning ehar er. Europ	rge (audi- pe 25p. C	o transiste commonwe	ors oniy), 1 ealth (Air) 1	5p extra pe Letter 65p	граіг.	el £1·69 (!	Min.).	1 w M/O 2% 4p. are E12 series. 1 watt 10%, 24p.	250k, 1M. LOG: 5k, 10k, 25k, 100k, 500
07A 13 k Pac	75p 2N 4929 Matcl king 13p per orde	ning ehar r. Europ Prices su	rge (audi pe 25p. C 1bject to	o transist Commonwe alteration	ors only), 1 ealth (Air) 1 n without p	5p extra pe Letter 65p rior notice.	граіг.			w M/O 2% 4p. are E12 series.	250k, 1M. LOG: 3k, 10k, 25k, 100k, 500 40p EACH.

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1N21 1N23	0.17	AFZ12 ASY26 ASY27	1-00 0-25 0-32	BYZII	0-85 0-32	OAZ211 OAZ222 OAZ223	0-82 0-45 0-45	ZS170 ZS271	0-10 0-18
1N85 1N253	0-88 0-50	ASY28 ASY28 ASY29	0-82 0-25 0-80	BYZ12 BYZ13	0-80 0-25	OAZ223 OAZ224 OAZ241	0.45	ZT21 ZT43	0-25 0-25
1N256 1N645	0-50 0-25 0-20	ASY36 ASY36 ASY50	0.25		1-00 0-62	OAZ241 OAZ242 OAZ244	0-22	ZTX107 ZTX108	0-15 0-12
1N725A 1N914 1N4007	0-20 0-07   0-20	ASY51 ASY51 ASY53	0-40	BYZ88C3V		OAZ246 OAZ290	0-23	ZT X 300	0.12
18113	0.15 0.18	ASY55 ASY62	0-20	C111 CR81/05	0.65	0C16 0C16T	0-50	ZTX 304 ZTX 500	0-25 0-16
18131	0.18	ASY86 ASZ21	0-38	CR81/40 C84B	0.45	0019	0.87	ZT X 503 ZT X 531	0-17 0-25
2G371 2G381	0-22	10700	0.75	CS10B DD000	3-18 0-15	0C20 0C22 0C23	0.50	INTEGR	TED
2G414 2G417	0-80 0-22	AUY10 AU101 BC107	1.50	DD003 DD006	0.15	0C24 0C25	0-80	CIRCUIT 7400	S 0-20
2N404 2N697	0.20	BC108 BC109	0-10 0-10	DD007 DD008	0-40	QC26	0.25	7401 7402	0-20
2N698 2N706	0-40	BC113 BC115	0-15 0-20	GD3 GD4	0-33	OC28 OC29	0-60 0-60	7403	0-20
2N706A 2N708	0-12	BC116 BC116A	0-25	GD5 GD8	0-88 0-25	OC30 OC35	0-40 0-50	7405	0-20
2N709 2N1091	0-63	BC121	0-25	GD12 GET102	0-05 0-80	OC36	0.60	$7407 \\ 7408$	0-30
2N1131 2N1132	0-25	BC122 BC125	0-20	GET103 GET113	0-22 0-20	0C41 0C42	0-25 0-30	7409	0-45
2N1302 2N1303	0-18	BC126 BC140	0-65 0-55	GET114 GET115	0-15	OC43 OC44 OC44M	0-40 0-17	7411	0-23 0-42
2N1304 2N1305	0-22	BC147 BC148	0.15	GET116 GET120	0-50 0-25	OC45	0.17 0.12	$   \begin{array}{c}     7412 \\     7413 \\     7416   \end{array} $	0-30
2N1306 2N1307	0-25	BC149 BC157	0.15	(4ET872 GET875	0-30	0C45M 0C46	0-18	7417	0-30 0-20
2N1308 2N2147	0-25	BC158 BC160	0-12 0-68	GET880 GET881 GET882	0-37 0-25 0-25	0C57 0C58 0C59	0.60 0.60 0.65	$7422 \\ 7423$	0-48 0-48
2N2148 2N2160	0.60	BC169 BCY31	0·13 0-35	GET882 GET885	0.25	0.066	0-60 0-50 0-12	7425 7427	0-48
2N2218 2N2219	0.20	BCY32 BCY33	0.55	GET885 GEX44 GEX45/1 GEX941	0.10	0C70 0C71 0C72	0.12	7428 7430	0-50 0-20
2N2369A 2N2444	0.15 1.99 0.28	BCY34 BCY38	0-30	GJ3M GJ4M	0-25	0073 0074	0-30	$\frac{7432}{7433}$	0-42 0-70
2N2613 2N2646 2N2904	0-45	110112/20	1.00	GJ5M GJ7M	0.25	1.0073	0-25	7437 7438	0-65
2N2904 2N2904 2N2906		BCY40 BCY42 BCY70	0.25	HG1005 118100A	0.50	0076 0077 0078	0.40	7440 7441AS	0-20 0-75
2N2907 2N2924	0-23	BCY71 BCZ10	0.20	MAT100 MAT101	0-25 0-30	0C79 0C81	0-22	5442 7450	0-75 0-20
2N2925 2N2926	0.15	BCZ11 BD121	0.50	MAT120 MAT121	0.25	0C81D 0C81M	0.20	7434 7458	0-20
2N3054 2N3055	0-50 0-75	BD123 BD124	0.80	MJE520 MJE2955	0-87	OC81DM + OC81Z	0.18	$7454 \\ 7460$	0-20
2N3702 2N3705	0-10 0-10	BDY11 BF115	1.62	MJ E3055 NKT128	0.87	1 OC82 OC821)	0-25 0-20	7470	0-80
2N3706 2N3707	0.28 0.12	BF117 BF167	0-50	NKT129 NKT211	0-30 0-25	0C83 0C84	0-25 0-25	7473 7474	0-40 0-40 0-55
2N3709 2N3710	0.10 0.10	BF173 BF181	0.25 0.85	NKT213 NKT214	0.25	0C114 0C122	0-39		0-45
2N3711 2N3819	0-10 0-85	BF184 BF185	0-20 0-20	NKT216 NKT217 NKT218	0-37 0-35	OC123 OC139	0-65	7482	0.87
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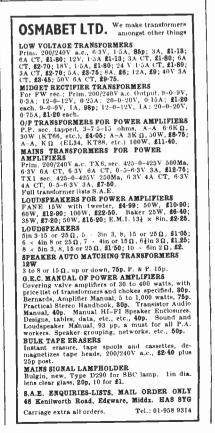
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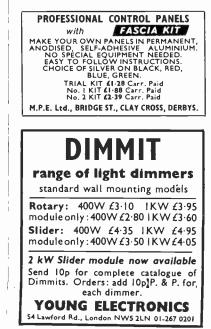
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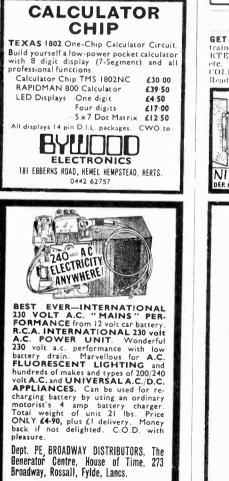


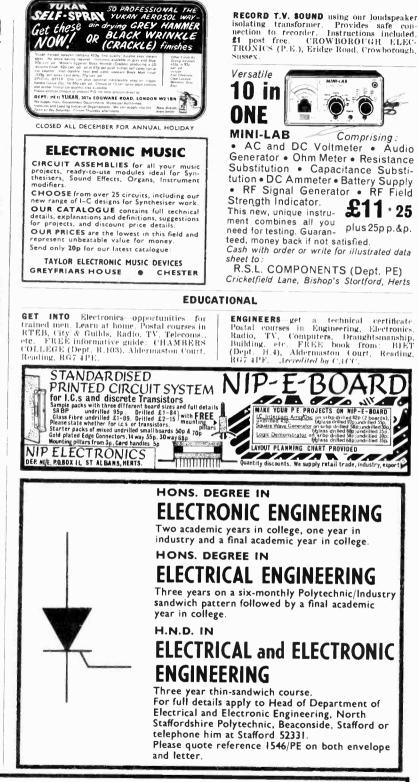


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number of copies of a printed circuit both simply and quickly.       15pf 4         Method: Spray copper laminate board with light-sensitive spray.       32pf 5         Cover with transparent film upon which circuit has been drawn.       39pf 6         Expose to light.       (No need to use ultra-violet.) Spray with developer,       68pf 1.5         Light-sensitive aerosol spray       100pf 2.2       100pf 2.2	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$
NEWER THAN NEW!!!!         VEROBOARD         FJH231/740           Fibre Glass Board pre-treated with light-sensitive lacquer enabling you to produce prototype printed circuits         2jin × lin × 0.15in 6p 3jin × 2jin × 0.15in 16p 5in × 2jin × 0.15in 16p 5in × 2jin × 0.15in 20p 5in × 200mm         FJH111/742 FJH171/745           150mm × 100mm	<ul> <li>OR capability</li> <li>OR capability</li> <li>I2p</li> <li>ON Dual 4-input NAND gate</li> <li>I2p</li> <li>Dual 4-input NAND buffer gate</li> <li>I2p</li> <li>Bual 4-input NAND buffer gate</li> <li>I2p</li> <li>Bual 4-input NAND buffer gate</li> <li>I2p</li> <li>Stimput AND/OR/NOT 2-level logic circuit</li> <li>I2p</li> <li>Edge-triggered Us flip-flop with two J, K and on J, K input</li> <li>Kinput</li> <li>Kinput</li> <li>Comparison</li> <li>Comparison</li></ul>
ELECTROLYTIC CAPACITORS       AXIAL LEADS       741         250µ1 25V       250µ1 25V       10p         2,000µ7 25V Rev.       25p       500µ1 25V       13p         1,000µ7 25V       35p       1,000µ1 25V       13p         10,000µ7 35V       35p       1,000µ1 25V       13p         10,000µ7 25V       35p       2,000µ1 25V       13p         10,000µ7 25V       35p       2,000µ1 25V       13p         10,000µ7 25V       35p       2,500µ1 50V       30p         60µ1 4200µ7 300V       30p       400µ1 40V       20p         10µ1 6V       2p       125µ1 4V       10p         10µ1 25V       4p       400µ 6.4V       12p         10µ1 25V       8p       320µ1 10V       3p         2µ1 25V       8p       320µ1 10V       3p         3µ1 10V       12pµ1 4V       12pµ1 4V       3p         3µ1 10V       12pµ1 4V       12pµ1 4V       3p         3µ1 10V       3p each       12pµ1 4V       8X728C         80µ1 10V       12pµ1 4V       12pµ1 4V       8X738C         80µ1 10V       12pµ1 4V       12pµ1 4V       8TX37/500R         80µ1 10V       12pµ1 4V       12pµ1 4V	A development of 709-OPA offering circuit simplifica- tion by employing internal frequency compensa- tion. Mounted in a miniature 8-lead moulded dual-in-line package, or TOS encapsulation



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