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U34 30 Blilicon PNP Alloy Trans. TO-5 BCY26 28302/4
U35 25 silicon Planar Transistori PNP TO-18 2N2906
U3n 25 Silicon Planar NPN Transistora TO-5 BFY50/51/52
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Code No's. mentioned above are given an a guide to the
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| Q6 | 5 OC 72 transiators |  |
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| Q8 | AC 126 tranilators PNP |  |
| Q9 | OC 81 type transistors |  |
| Q10 | 7 OC 71 tgpe tranaiators |  |
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| Q12 | 3 AP 116 type tranalators | 0.50 |
| Q13 | 3 AP 117 type transistors | 0.50 |
| Q14 | 3 OC 171 H.F. type transintors |  |
| Q15 | 7 2N2926 Stl. Epozy tranaistors mired colours |  |
| Q16 | 2 GET880 low nolse Germaniun transistors |  |
| Q17 | 5 NPN $2 \times 8$ P. $141 \pm 3 \times 8$ S. 140 |  |
| Q18 | 4 MADT'8 $2 \times$ MAT 100 \& $2 \times$ MAT |  |
| Q19 | 3 MADT'S $2 \times$ MAT $101 \& 1 \times$ MAT 121 |  |
| Q20 | 4 OC 14 Germanium transistors A.F. |  |
| Q21 | 4 AC127 NPN Germanium trangistors | 0.50 |
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| Q24 | 8 OA 81 diodes |  |
| Q25 | 15 IN914 Biticon diodes 75 PIV 75 mA |  |
|  | 8 0A9s Germanlum diodes sub-min IN69 |  |
| Q27 | 2 10A PIV glicon rectliers 18425 | 0.50 |
| Q28 | 2 Bilicon power rectifiers BYZ 13 |  |
| Q29 | 4 Silicon transistors $2 \times 2$ N $^{2} 96$, $1 \times 2$ N697, $1 \times 2$ N68 $^{2}$ | $0.50 .$ |
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| Q31 | 6 sllicon wwitch transistora $2 N 708$ NPN | $0.50$ |
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| Q37 | 3 2N3053 NPN Silicon transietors |  |
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| BP709-72709 | (10 | ${ }^{34}$ D | 80 p |
| BP 709P- $\mu \mathrm{A} 709 \mathrm{C}$ |  | 34 D | 80 p |
| BP 710-72710 | 44p | 48D | 40 p |
| BP $711-\mu \mathrm{A} 711$ | 45p | 48D | 40 p |
| BP 741-72741 | 75 p | 60 p | 50 p |
| $\mu \mathrm{A} 703 \mathrm{C}-\mu \mathrm{A} 703 \mathrm{C}$ | 28D | ${ }^{28} \mathrm{D}$ | 24p |
| TAA $283-$ | 70p | 80p | ${ }^{65 p}$ |
| TAA $293-$ | ${ }_{90 \mathrm{p}}$ | 750 | 70 D |
| TAA 850 | 170p | 158p | 150 p |
| B.G.S. RA1000 8.68 |  |  |  |

Typ
-

| modbl | cmes | 08116 |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Anode roltese( (Jde) | 17 mma | 178mim | - |  |
| Cathode Currat (mA) | 2.3 | 14 | 8 |  |
| Numerical Height (mm) <br> Tube Height (mm) | ${ }_{4}^{18}$ | 13 <br> 3 <br> 1 | 22 |  |
| Tabe Dinater (mm) | 19 | 13 | 12 wide | atio |
| L.C. Divere Rec. | ${ }_{\substack{\text { Britio } \\ 141}}^{\text {or }}$ | $\underset{\substack{\text { Bratia } \\ 19}}{19}$ | ${ }^{\text {br47 }}$ |  |
| $\overline{\text { Price EACH }}$ | 81.70 | 31.88 | 2 |  |


| RTL MICROLOGIC CERCUTS |  |  |  |
| :---: | :---: | :---: | :---: |
|  | Price each |  |  |
| Epozy TO-5 case | 1-24 | 26-99 | 100 dp |
| uL900 Buffer | 85 p | 38p | 87 |
| uL914 Dual 21/p gate | 85p | 88p | 87p |
| uL923 J-K fip-flop | 500 | 47p | 45 |
| Date and Circuit Price 7p. | $\mathrm{BO}_{0}$ | klet for | IC' |

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| :---: | :---: | :---: | :---: |
| ${ }_{\text {SN7 } 773}$ | ${ }_{0}^{0.37}$ | ${ }_{0} \cdot 35$ | 0.32 |
| 8N7474 | - 0.37 | - 0.35 | - |
| ${ }_{\text {8NT7476 }}$ | - | - | ${ }_{0}^{0.38}$ |
| 8N7480 | 0.87 | 0.64 | 0.58 |
| $\mathrm{BNT}^{\text {P81 }}$ | 81.20 | 21.15 | 81.10 |
| 8N7482 | 0.87 |  | 0.85 |
| $8 \mathrm{B7} 483$ | 21.10 | ${ }^{21} .05$ | 0.95 |
|  | ¢1.00 | - 3.50 |  |
| ${ }_{\text {BN7 } 7486}$ |  | 0.81 | 0.80 |
| ENT499 | ${ }_{56} 5.50$ | ${ }^{255} 0.84$ | 85.00 |
| ${ }^{\text {SNN }} 7490$ |  |  | - 0.80 |
| ${ }^{8 N T} 491$ | 0 | ${ }_{0}^{0.65}$ | 8 |
| ${ }_{8 N \mathrm{SN} 743}$ | 0.67 | ${ }_{0} .84$ | 58 |
| 8N7494 | $0 \cdot 77$ | 0.74 | 88 |
| 8N7495 | 0.77 |  | 88 |
| EN7496 | 0.87 | 0.84 |  |
|  | ${ }_{21}^{21.65}$ | 81.80 |  |
| 8N74104 | 0.87 | 0.94 | 88 |
| 8N74107 | 0.40 | 0.38 | 80 |
| ${ }^{8 N 874111}$ | 81.25 | ${ }_{11} 1.15$ | ${ }_{81} 1.10$ |
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| - ${ }^{\text {8N74419 }}$ | - 11.35 | ${ }^{21.25}$ | ${ }^{81.10} 0$ |
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| 8N74153 | ${ }_{81}^{81} 80$ | ${ }_{81} 1.70$ |  |
| SN74155 | 81.40 | 81.30 |  |
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| BP932 | 18p | 12p | 11p |
| BP989 | 18 | 12. | 11 |
| BP935 | 18p | 12p | 11p |
| BP936 | 18. | 129 | 11p |
| 9784 | - 18 | 129 | 11D |
| BP945 | 25p | 240 | 2\%9 |
| RP946 | 12p | 11p | 10 p |
| BP948 | 25p | 24D | 28D |
| BP951 | 65 p | 60 p | 659 |
| BP962 | 12p | 11p | 109 |
| BP9093 | 40 D | 38p | 85 p |
| BP9094 | 40 p | 38 p | 86 |
| BP9097 | 40p | 88p | 35 |
| BP9099 | 40p | 38p | 85p |

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## THE HY41



The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically $0.05 \%$ at 1 KHz into 8 ohms! and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 retains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for $\mathrm{Hi}-\mathrm{F}_{\mathrm{i}}$ but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.
OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts
R.M.S. continuous.

LOAD IMPEDANCE: 4-16 ohms.
INPUT IMPEDANCE: 30 K ohms at 1 KHz .
VOLTAGE GAIN: 30 db at 1 KHz
TOTAL HARMONICDISTORTION: less than 0.15\% (typical 0.05\%)
at 1 KHz .
FREQUENCY RESPONSE: $5 \mathrm{~Hz}-50 \mathrm{KHz}+1 \mathrm{db}$.
SUPPLY VOLTAGE: +22.5 voits D.C.
SUPPLY CURRENT: $\overline{0}, 8$ amps maximum.
PFICE : inc. comprehensive manual, P.C. board, five extra components and P. \& P.:MONO: $£ 4.90$ STEREO: $£ 9.80$

## UNIQUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier ta contain all feedback and equalization networks within an integrated preamplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use.

Internally the HY5 provides equalization for almost every conceivable input, the desired function is achieved by use of a multi-way switch or by direct interconnection

Two dissinceive features of the HY 5 are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from $16-25$ Volts and a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

## INPUTS

Magnetic Pick-up (within $\pm \mathbf{1 d b}$ RIAA curve) $2 \mathrm{mV} .47 \mathrm{~K} \Omega$
Tape Replay lexternal components to suit head). $4 \mathrm{mV}, 47 \mathrm{~K} \Omega$
Microphone (flar) $10 \mathrm{mV} .47 \mathrm{~K} \Omega$
Ceramic Pick-up lequalized and compen-
satablel $20-2000 \mathrm{mV}$. variable.
Tuner (flat) 250 mV . $100 \mathrm{~K} \Omega$
Auxiliary 1250 mV . $47 \mathrm{~K} \Omega$
Auxiliary $22-20 \mathrm{mV} .100 \mathrm{~K} \Omega$

## OUTPUTS

Main Preamp output 500 mV .
Direct tape output 120 mV .
ACTIVE TONE CONTROLS (Bexendall)
Treble + 12db.
Bass +12 db .
INTERNAL STABILIZATION
Enables the HY5 to share an unregulated
supply with the Power Amplifier.
SUPPLY VOLTAGE
16-25 volts
PRICE:
MONO: $£ 3.60$

STEREO: $£ 7.20$

## POWER SUPPLY PSU45

The versatile P.S.U. 45 is designed to supply your HY41's + HY5's in stereo or mono format.

Specification
Input: 200-240 Volts
Output: +22.5 Volts at 2 amps
Overall Dimensions: L. $7^{\prime \prime}$, D. 3.8'; H. 3.1"
PRICE: $£ 4.50$ inc. $P . \& P$.

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## STEREO <br> IC DECODER

HIGH PERFORMANCE PHASE LOCKED LOOP (as in 'W.W.' July '72) MOTOROLA MCIBIOP

## EX STOCK DELIVERY

Specn. Separation: $40 \mathrm{~dB} 50 \mathrm{~Hz}-15 \mathrm{kHz}$. Distortion: $0.3 \%$. I/P level: 560 mV rms. O/P level: 485 mV rms per channel. Input impedance: 50 k . Power requirements : $8-12 V @ 16 \mathrm{~mA}$. Witl drive up to 75 mA stereo on' lamp or LED.
KIT COMPRISES FIBREGLASS PCB
(Printed and tinned), Resistors, I.C., Capacitors, Preset Potm, \& Instructions. Only $£ \mathbf{3} \cdot \mathbf{5 0}$ post free.
LIGHT EMITTING DIODE (Red)
Suitable as stereo 'on' indicator. For above with panel mounting clip and instructions. Only 35p + p.p.
MCI3IOP only E2.77 + p.p. $6 p$

Fi-Comp Electronics
BURTON ROAD, EGGINTON, DERBY DE6 6GY

## ＝ 5 WherecanIgetexclusive units at lowest prices？＂．

 Eachatimulcs！

Right Mic sackats on front panel．Dual recording level meter，and Left and Right volume controls．Built－in pre－amp．Tape speed $3 \frac{2}{2}$ ips $(9.5 \mathrm{cms})$ ．Frequency response：playbach $30-10.000 \mathrm{~Hz}$ ：recording／playback $30-8,000 \mathrm{~Hz}$ ．Line output：fully vatiable 0.500 mV ．The RP．1000ST incorporates all the features you＇d expect on a top quality Cartridge Tape Deck． Site $16^{"}$ wids． $4^{\prime \prime}$ high， $9^{\prime \prime}$ deep．Cabinet in walnut．Including connecting leads． REALISTIC 30 WATI SIEREO
AMPLIFIER MODELSA．500 AS RVi8Ned tures you ve vert wanted－for under－ Cha．Saving over f10－00 on the nor volume and balance controls．Headphone socket on froni panel．Push－buttor input controls－magnetic phono（high／low）suner，aux，mano，manitor． Loudness push－bution control for perfect sound at low output levels Left and right push－button on／off switches for speakers．Notse filtering 835 ad rape monitoring facilities．Iwo auxiliary AC outlets．Frequency response 20 C 55 $20.000 \mathrm{~Hz} \pm 1 \mathrm{db}$ at full power． 15 watts rms per channel．Walnut cabinet with satin aluminium trims．Inputs phono 2.5 mV and 5 mV R1AA；tuner／aux 250 mV ．Hum and noise：phono－50db；tunerfaur－65db．How＇s that for a soecification！

## \＆

8．TRACK STEREO CARTRIOG TAPE OECK MOOEL RP－1000SY The popular Lear Jes type record ing unit is the heart of the fantastic RP－1000ST．which has full record and alayback facilities．Automatic ack change with manual over－ride．
 The A．3000 looks as good as it sounds！ Giving you big，performance this superb udio amplifier hàs a full range of facilities on the front and rear panels．On the front－ Il the cortrols you＇re evef tikely to need lus a headphone socket．On the rear signal inputs speaker outputs and a line luse for circuit protection．
Specifications： 18 watts rms per channel 2to 8 ohms．Frequency response $20-35,000$ $\mathrm{Hz}( \pm 2 \mathrm{db}$ ）Inputs Magnetic．Ceramic Tuner．Jape，Aux．Tape Play．Size $45 \mathrm{~mm} \times 300 \mathrm{~mm} \times 130 \mathrm{~mm}$

## WATTS＂ Automatic Record Cleane keeps your records clean as they play．E1． 16 Watts Disc Pieener．Keeps new records like new 10r perfect record repr

## R． 307 TRANSISTOR－

 IZEO STEREO PRE－AMPLIFIER
## that could anly reprod

 accept signals from maving－magnet cat ridges！The R． 307 steps up signals from between $5 \cdot 20 \mathrm{mV}$ to $200-800 \mathrm{mV}$ ．Input 5.20 mV ．Equalisation：RIAA．Ourputt： 200.800 mV flat．Frequency range： 20 $22,000 \mathrm{~Hz}$ ．Dimensions： 3 ，$\times 1 / \times 22$
Supply： 240 VAC．RUC PRICE 4.92

## （6）

15．FOOT STEREO HEADPHONE
EXTENSION CORD R， 362 3 Ex Fitted with heavy duly and a maching stereo sockel a the other．ROC PRICE E1． 30 STEREO HEADPHONE＂$Y$＂＊ADAPTOR R． 361 Enables you to use two sets of stereo headphones from a single sock sockets．ROC PRICE E1－30

OLSON AM－39 40－WAYT STEREO
AMPLIFIER An ideal unit for your new It is more than $\mathbf{~} 10.00$ below the normal retail
prical Making the AM－ 395
 to own a transcription cartidpe for the price of＇a ceramis！Is specially de． signed to match top quality tone arms．and to get the very best from your hi－fi amplifier． 0.7 mil diamond stylus．Output： 7 mV pe Channat balanency range： $20-21.000 \mathrm{~Hz}$ tion：28d8．Compliance： $12 \times 10.6 \mathrm{~cm} /$ dyn RDC PRICE EF．37


RR． 446 3－WAY MATCHED SPEAKERS pocket At ondy 516.40 a pair they are real value for you Each handle 16 watts rms（ 8 watts rms each）．Each loudspeaker contains a large dual cone base unit plus a separate twenter．Frequency range： 40 to $19,000 \mathrm{~Hz}$ ．Size $14^{\circ}$ high

EAGLE LC． 05 STEREO For tabulous reproduc．
lion at a very low price．

0.7 mil till

0．1． 6 mV per stylus．
30－18，000 Hz ．Channel batance：range Channel separation：20dB．Recommended stylus piessure： 2.4 grams．Compliance
$9 \times 10.6 \mathrm{~cm} / \mathrm{dyne}$ ．ROC PRICE f 4.75

##  6 最首电 Normal Price $\mathbf{C 1 5 . 7 0}$

## OLSON AM－357 4－WATt Stered amplifier

Here s marvellous value for somsone just slating to set them selves up in audio！At only $[10,50$ ，you get a fine amplifies in a front panel．It incorporates separate tone and volume contrals for each channel．Inputs are provided for turntable（ceramic car－ tridge），funer and tape deck or recorder．Frequency response： 70 $20.000 \mathrm{~Hz}+3 \mathrm{~d} 8$ ．Dutput： 2 watis r．m．s．per channel into 8 ohms Inputs：phono 80 mV ：tuner／auk 80 mV ．Size $8^{*}$ wide． $21^{\circ}$ high

## R

CRYSLER ＇LIVING AUDIO＇ SPEAKER CE－5b

This nigh quality speak． er has its own built－in 3．way sound response switc
 hi－fi．natural or mood musir listening lts beautitul heavy oiled watni caing． incorporates two separate speaker units an 8 wool mid－range with $2^{\circ}$ concentric tweeter．Power kandling

## capacity： 25 watts r．m．s into 8 ohms．Overall frequenty res－

 ponse： $35-20,000 \mathrm{~Hz}$ ．Cabinet size： $10 \frac{10}{\mathbf{N}^{\prime \prime}} \times$
### 827.65

## ©

PALACE AM／FM／MPX STEREO TUNER AMPLIFIER SSA－IG This is one of the lowest priced stero tuner amplifiers on the market．It covers the full range of both AM and FM broadcas frequencies．And when you re switched to GM，an indicator hights up when stereo signal is received－that＇s the time to switch to＂Stera0．1 The SSA－16 has all the facilities You＇d ex pect to find on tunars costing twice as much－separate vol ume．thass，treble，balance and tuning controls．Selector switch for tape．phono．AM．FM．stereo．Jack socket on front panel for stere headphones．Frequency range：$F M 88-108 \mathrm{MHz}_{2}$ ． AM 535.1605 kHz ．Frequency response： $50.10 .000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ ， Power output： 4 watts lotal musir power into two 8 ohm speakers．Size： $16^{\prime \prime}$ wide， $4 \frac{1^{*}}{}{ }^{*}$ high， $8^{\prime \prime}$ deep．

compact unit measuring only $5 \frac{1}{2}$ wide． 12
high and $6 \frac{1}{7}$ deep．It contains its own mains
power supply，and has a ganged volume control and separate trense $40-17000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ ；putput $3-5$ watls music power per channe！into 8 ohms；input．phono． 600 mV ：signal－to－noise ratio better than 45 dB ．

## （1）

## 

dison am－372 16－watt stereo amplifier $P$ here sa really good amplifier at a realy down－to－sanh price the AM－372 will do for you－reproduce signals from seramic or crystal eantridges．AM and FM tuners，and tape tecorders． And it gives you outputs lor two sets of speakers，headphones and tape recorders．Frequency response is 30 to $20.000 \mathrm{~Hz} \pm$ 3dB．Output 8 watts f．m．s．per channel music power into 8 ohm speakers．Phond input 200 mV ．
prical Making the AM－ 395

## 294， 11 buys．It takes in signals from magnetic of coramic pick－up，

 uts for taping and for headphones．There are separate bass and treble controls．separate Left and Right channel volume controls．And a loudness switch for boosting the bass and Trebte notes when listening at low output levels．Frequency
 specification reads well－sounds aven better

Return－ol－post mail order vice．Orders over $\mathbf{£ 1 0}$
post free（UK only）． Add $25 p$ for $p$ \＆$p$ to orders under f 10 ． HP terms vailable for callers

## BSR LATEST SUPERSLIM STEREO \& MONO <br> Plays $12^{\circ} 10^{\circ}$ or $7^{*}$ record Auto or Manual. ${ }^{\text {a }}$ high quality unit backed by ESR guarantee. AC 200/250\%. Size $13 \mathrm{f} \times 11 \mathrm{tin}$. Above motor board 32 in . with STEREO and MONO XTAL MONO-COMPATIBLE Plays all records <br>  E8.75 Post 25 RCS DE-LUXE 3 WATT AMPLIFIER, Ready mede, tested. Printed circuit. 3 watts output. Tone and volume knobs, and high performance loudapeaker. Double wound fully isolated mains transtormer. A.C. mains 200/250V Response $50-12,000 \mathrm{cps}$. Sensitivity 300 mV . $\quad \mathbf{~} \boldsymbol{4}_{25 \mathrm{p}}^{\text {Post }}$

GARRARD DISCO DECK € 10 Post Free 4 speeds. Plags all sizes of records. pole bervy duty
motor. gin. steel turntable. Plug in
 head complete with stereofmono cartridge. Adjustable stylus pressure. Auto-
stop. Brown and Cream faish. AC mains $110 / 240 \mathrm{p}$. Base plate size $12 i n \times 8 i n$. Operating area $14 i n . \times 12$ in. Above motor board sin, below motor board 2sin Ideal for Home Hi-Fi or Discotheque.
GARRARD ADTOCHANGERS with Sonotone Cartridges Btereo Dlamond and Mono Sapphire. Model $1025 ~ £ 10$
Model 3500 Stereo and Mono Autochanger £14. Post 25 p

BSR JUNIOR

## SINGLE PLAYER

Heavy duty 4-speed motor with separate pick-up arm
LP/78 turnover mono $£ 4.50$
 cartridge

EI-FI PICE DP CARTRIDGES. Diamond LP/Stereo Stereo/Mono 9TA £2.50; GP94 £2.50; GP93 £2
Sapphire Mono GP91 £1-50; Powerpoint LP/78 60

GARRARD DECCADEC SP25 Mk. II RECORD PLAYER

## Single play Stereo

Mono Deram
transcription
head and arm
Four speeds
Anti-rumble fle Bias compensation. Laboratory motor.


METAL PLINTH AND PLASTIC COVEI Cut out for most Garrard or B.S.R. Most will play with cover in positio Latest Design. Covered in blac $12 \frac{1}{2} \times 14 \frac{1}{2} \times 7 \frac{1}{3}$ in high
ALSO AVAILABLE IN SOLID NATURAL MAHOGA WAX POLISHED FINISH AT SAME PRICE
COAXIAL PLUG 6p. PANEL SOCKETS 6p. LINE 18p. OUTLET BOXES. SURFACE OR FLUSH 25 p
BALANCED TWIA FEEDERS 80 ohms or 300 ohms. 5 p yd Chrome Lead Socket 45 p . Phono Plugs 5 p . Phono Socket 5 p. JACK PLUGS Std. Chrome 15p; 3.5 mm Chrome 14p. DIN SOCKETS Chassia 3-pin 10p; 5-pin 10p. DIN SOCKETS Lead 3-pin 18p; 5-pin 15p. DIN PLUGS 3-pin 18p; 5-pin 25p.
$\qquad$
BLANK ALUMINTUM CHASSIS. 18 g.w.g. $2 \frac{1}{2} \mathrm{in}$. sides $6 \times 4 \mathrm{in} .45 \mathrm{p} ; 8 \times 6 \mathrm{in} .53 \mathrm{p} ; 10 \times 7 \mathrm{hn} .65 \mathrm{p} ; 19 \times 8 \mathrm{in} .85 \mathrm{p}$; $1.4 \times 91 \mathrm{n}$. 90p: $16 \times 8 \mathrm{in} .90 \mathrm{p} ; 12 \times 31 \mathrm{n} .50 \mathrm{p} ; 16 \times 10 \mathrm{in}$. $\mathrm{f1}$. $14 \times 3 \mathrm{in} .16 \mathrm{p} ; 10 \times 7 \mathrm{in} 19 \mathrm{p} ; 12 \times 5 \mathrm{in} 20 \mathrm{p} ; 12 \times 8 \mathrm{in} .15 \mathrm{p}$; $16 \times 6 \mathrm{in}$. $28 \mathrm{p} ; 14 \times 9 \mathrm{in}$. $34 \mathrm{p} ; 12 \times 12 \mathrm{in} .40 \mathrm{p}$; $16 \times 10 \mathrm{n}$.. 50 p ;

## RADIO BOOKS (Post Paid

Handbook of transistor equivalents
Radio valve guide, Book 1, 2, 3, 4, 5 (each) Master colour code folder
Boys' book of crystal sets and simple circuits Migh fidelity loudspeaker enclosures Practical Stereo bandbook Transistor superhet receivers Coil design and construction manual Radio, TV and electronics data book Transistor circuits manual, No. 4 Practical transistor audio amplifier Transiator subminiature receivers Manual of transiator audio amplifiers Book 2 Resistor colour code disc caleulator TV fault Inding
R.C.S. STABILISED POWER PACK KITS

All parts and inatructions with Zener Diode, Printed Circuis, Bridge Rectiters and Double Woand Mains Transformer
inpat $200 / 240 \mathrm{~V}$ a.c. Outpat voltages a vailable 6 or $\theta$ or 12 ${ }^{\text {or }} 15$ or 18 or 20 ave dic. at 10 amA or less Please state voltage reqoired. Details S.A.E. Size $3 \downarrow \times 1 \nmid \times 1$ inin. $\leq 2$
R.C.S. GENERAL PURPOSE TRANSISTOR PRE-AMPLIFIER BRITISH MADE
Ideal for Mike, Tape, P.U., Guitar, etc. Can be nsed with Battery $9-12 \mathrm{~F}$. or a.T. line $200-300 \mathrm{~F}$. D.C. Operation. Size For use with valve or transistor equipment. 90 Post Full inatructions supplied. Details S.A.E. 90p 10

N
NEW TUBULAR ELECTROL
 $8 / 450 \mathrm{~V}$. $14 \mathrm{p} \quad 1000 / 25 \mathrm{~V} \quad 35 \mathrm{p}$
$16 / 450 \mathrm{~V}$
$32 / 450 \mathrm{~V}$
$32 / 450 \mathrm{~V}$
$25 / 25 \mathrm{~V}$
$32 / 45 \mathrm{~V}$
$25 / 25 \mathrm{~V}$

| $25 / 25$ | $\cdots$ | 10 p | $8+16 / 450 \mathrm{~V}$ | 20 p | $350+50 / 325 \mathrm{~V}$ |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $50 / 50 \mathrm{~V}$ | $\cdots$ | 10 p | $16+18 / 450 \mathrm{~V}$ | 25 p | $32+32+30 / 050 \mathrm{~V}$ | $\begin{array}{lllll}50 / 50 \mathrm{~V} & 10 \mathrm{p} & 16+18 / 450 \mathrm{~V} 25 \mathrm{p} & 32+32+32 / 350 \mathrm{~V} 43 \mathrm{p} \\ 100 / 25 \mathrm{~V} . . & 10 \mathrm{p} & 32+32 / 350 \mathrm{~V} 25 \mathrm{p} & 100+50+50 / 350 \mathrm{~V} 48 \mathrm{p}\end{array}$ LOW VOLTAGE ELECTROLYTICS

$1,2,4,5,8,16,25,30,50,100,200 \mathrm{mF} 15 \mathrm{~V} 10 \mathrm{p}$ $500 \mathrm{mF} 12 \mathrm{~V} 15 \mathrm{p} ; 25 \mathrm{~V} 20 \mathrm{p} ; 50 \mathrm{~V} 30 \mathrm{p}$.
$1000 \mathrm{mF} 12 \mathrm{~V} 17 \mathrm{p} ; 25 \mathrm{~V} 35 \mathrm{p} ; 50 \mathrm{~V} 47 \mathrm{p} ; 100 \mathrm{~V} 70 \mathrm{p}$.
2000 mF 8V $25 \mathrm{p} ; 25 \mathrm{~V} 42 \mathrm{p} ; 50 \mathrm{~V} 57 \mathrm{p}$ $2000 \mathrm{mF} 6 \mathrm{~V} 25 \mathrm{p}: 25 \mathrm{~V} 42 \mathrm{p}: 50 \mathrm{~V} 57 \mathrm{p}$
2500 mF 50 V 62 p


CERAMIC, 1 pF to $0.01 \mathrm{mF}, 4 \mathrm{p}$. Silver Mica 2 to 5000 pF . 4 p . PAPER $350 \mathrm{~V}-0.14 \mathrm{p}, 0.513 \mathrm{p} ; 1 \mathrm{mF} 15 \mathrm{p}$; 2 mF 150 V 15 p . S00V-0.001 to $0.054 \mathrm{p} ; 10.15 p ; 00258 p ; 0.4725 \mathrm{p}$. $2,200 \mathrm{pF} 10 \mathrm{p}: 2,700-5,600 \mathrm{pF} 20 \mathrm{p} ; 6.800 \mathrm{pF}-0.01$ prd 30 p . $880-$ TWIN GANG. "0-0" $208 \mathrm{pF}+176 \mathrm{pF}, 65 \mathrm{p}$; Slow motion drive $385 \mathrm{pF}+385 \mathrm{pF}$ with $25 \mathrm{pF}+25 \mathrm{pF}, 50 \mathrm{p}$; 500 pF slow motion, atandard 45 p ; small 3 -gang 500 pF . $£ 1$-60.
SHORT WAVE SINGLE. SHORT WAVE SINGLE. $10 \mathrm{pF}, 30 \mathrm{p}, 25 \mathrm{pF}, 55 \mathrm{p}, 50 \mathrm{pF}, 55 \mathrm{p}$.

SHORT WA Gangable Tunlng Condensers.
Values up to 100 p .
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Section Couplers supplied FREE with two or eac
NEON PANEL INDICATORS 250 V AC/DC Red or RESISTORS

 Witios io Previd RESISTORS 5 watk 10 wegat ${ }^{4 p}$ 10 o /ms to 100 K 10p eaoh; $2 \frac{1}{1}$ watt, 1 ohm to 89 ohms 10 p .

## MAINS TRANSFORMERS

$250-0-25080 \mathrm{~mA} .6 .3$ v. 4 mmp .
$250-0-25080 \mathrm{~mA} .8 .3$ จ. 3.5 a. 6.3 च. 1 a, or 5 р. 2 a .12 .50
 MINIATURE $200 \mathrm{v}, 20 \mathrm{~mA}, 6.3 \mathrm{v}, 1$ a $21 \times 21 \times 2$. MIDGET 220 ₹. 45 mA , $8.3 \mathrm{v}, 2 \mathrm{a} .21 \times 2 \mathrm{z} \times 2 \mathrm{in}$. MINI-MAINS $20 \mathrm{v}, 100 \mathrm{~mA}$. $1 \ell \times 1 i \times 1 \| \mathrm{in}$. HEATER TRANS. $6.3 \mathrm{\nabla} .3 \mathrm{~s}$
Ditto te pped sec. 1.4 v. $2,3,4,5,6.3 \mathrm{v} .11$ amp. t $2 \mathrm{amp}, 3,4,5,6,8,9,10,12,15,18,24$ and 30 $1 \mathrm{amp} .6,8,10,12,16,18,20,24,30,36,40,48,60$ amp., $8,8,10,12,16,18,20,24,30,36,40,48,60$ $\mathrm{amp} ., 8,8,10,12,18,18,20,24,30,38,40,48,60$ $3 \mathrm{amp} .5,8$ and 13 V
$3 \mathrm{amp} .3,8,8,10,13$ and $5-0-5 V$
$5 \mathrm{amp} .3,5,8,10,13$ and $5-0-5 V$
ADTO TRANSFORMERS. 115 v to 230 v or 230 v to CHARGER TRANSFORM: 750 w . £10; 1000 w . £14 or 6 or 12 F ., 1 t FULL WAVE BRIDGE CHARGER RECTIFIERS


## MAINS ISOLATING TRANSFORMER

Primary 0-110-240p. Secondary 0-240\%, 3 ampg 720
watts. Intulated terminals. Varnish impregate enclosed in ateel case with fixing feet. fig Carrisge enclosed in ateel case with fixing feet.
Famons make. BARGALI OFFER

\section*{WETRAD P50 TNANSISTOR COILS} RA2W Ferrite Aerial. I.F. P50/2CC $470 \mathrm{kc} / \mathrm{s}$ | Osc. P50/1AC |  |
| :--- | :--- |
| I.F. P50/20C 470 kc/a....36p | Driver Trans. LFDT4 |
| Srd I.F. P50/3CC | Printed Citcuit. PCA1 | 72p ${ }^{73 \mathrm{p}}$ Spare Cores. P51/1 or P51/2 ..........36p | 36p | $\begin{array}{c}\text { J.B. Tuning Gang } \\ \text { Weyrad Booklet }\end{array}$. |
| :---: | :---: | P50/3V $\begin{array}{ll}36 p & \text { Weyrad Bookles } \\ 36 \mathrm{p} & \\ \text { OpT1 }\end{array}$

Wullard Ferrite Rod $8 \times \operatorname{lin}$. 20p. $6 \times$ in. 20 p

VOLUME CONTROLS ${ }^{80}$ obm Coax 4p yd. Long spindles. Midget Size BRITISH AERIALITE 5 K ohms to 2 Meg LOG or AERAXIAL-AIR SPACED LIN. L/S 15 p . D.P. 25p. $40 \mathrm{yd} . \varepsilon 1 \cdot 40 ; 60 \mathrm{yd} .22$. | STEREO L/S 55p. D.P. 75p. | FRINGE LOW LOSS |
| :---: | :---: | :---: | :---: |
| Edge 5K. S.P. Transistor 25 p . | Ideal 625 and colour |

1 finch DIAMETER WAVECHANGE SWITCHES. 25 p .
2 p. 2-wag, or 2 p. 6 -way, or 3 p. 4 -way 25 p each 1 p.1 2-w89, or 4 p. 2-way, or 4 p. 3 -way $25 p$.
TOGGLE SWITCHES, sp. $14 p$; dp. 18 p ; dp. dk. 23 p .

E.M.I. $13 \frac{1}{2} \times 8 \mathrm{in}$. SPEAKER SALE!
With twin tweeters.
And crossover. 10
watt. State 3 or 8 or 150 hm . As illugtrated. Post 25 p With Gared tweeter cone and ceramic magret. 10 watt. Bass res. $45-60 \mathrm{cp}$ Blax res. $450-80 \mathrm{cps}$.
Flans. t2.25 State 3 or 8 or 15 ohm. Post 25
 GOODMANS $6 \frac{1}{2}$ in. HI-Ft WOOFER 8 ohm. 10 watt. Large ceramm Special Cambric cone surround.
Frequency response $30-12.000$ cps. Ideal P.A. Columns. Hi-Fi Enclosure Systems, etc

## ELAC CONE TWEETER

The moving coil diaphragm gives a good radiation pattern to the bigher frequencies and a smooth extension of total reaponse from $1,000 \mathrm{eps}$ to $18,000 \mathrm{cps}$. Size $3 \frac{1}{2} \times$ 15 ohm models. $\mathbb{E} .90$ Post 10 p

## GOODMANS

 8 in WOOFER8 ohm 12 watt. Deep cone. resonance 35 cps . Frequency response $30-8,000 \mathrm{cps}$.

## c4.50



SPECIAL OFFER! 80 ohm. $2 \mathrm{ikin} 2 \exists \mathrm{in} .35$ ohm, $2 \mathrm{in} \cdot ; 3 \mathrm{in}$.
 $15 \mathrm{ohm}, 3$ inin, dia.: $6 \times 4$ in; ${ }^{2} \times 4 \mathrm{in}$;
LODDSPEAKERS P.M. 3 OHMS. 7 (4in. £1-25; 6 in. £1-50; $8 \times 5 \mathrm{in} . £ 1.60 ; 8 \times 21 \mathrm{in}$. $21.50 ; 8 \mathrm{in}$. $81.75 ; 10 \times 6 \mathrm{in}$. $£ 1.90$. RICHARD ALLAN TWIN CONE LODDSPEAKERS 8 in . dia. 4 watt; 10 in . dia. 5 watt; 12 in. dia. 8 watt, 3 in. dia, 4 watt; 10 in. dia. 5 watt; 12 in. dia. 8 we 15 ohm models 22 each. Post $15 p$.
SPEAKER COVERING MATERIALS. Samples Large S.A.E. Horn Tweeters $2-16 \mathrm{~K} / \mathrm{s}, 10 \mathrm{~W} 8$ ohm or 15 ohm £ $1-50$. De Lure Horn Tweeters $2-18 \mathrm{Kc} / \mathrm{s}$, $15 \mathrm{~W}, 15 \mathrm{ohm} 23$. TWO-WAY 3,000 c.p.s. CROSSOVERS 3 or 8 or 15 ohm 95 p .

TWO-WAY CROSSOVER NETWORK $3,000 \mathrm{c} / \mathrm{s}$ With varlable tweeter attenustor giving accurate high/low frequency balance. Mounted on panel $5 \frac{1}{n}$ in. $\times 4 i n$. with
 VALVE OUTPDT TRANS. 28p. MIKE TRANS. $50: 1$ 25p 5 WATT MULTI RATIO, 3,8 and 15 ohms 80 p .
HEAVY DUTY PUSH-PULL OUTPUT TRANSFORMERS 50 watt …........ $£ 9.50 \quad 100$ watt $\quad . . . . . .$.

## \$TEREO/MONO HEADPHONES <br>  <br> Model 6000 with new type slider volume controls. Stereo/Mono ${ }^{8}$ Witch. 8 ohms. £10. volume controls. 8 ohm. $£ 7$. Budget Model. 8 ohm, $£ 3-25$. Stereo Stethoscope 8 ohm £1.98. HEADPHONE JONCTION BOX with awitch $£ 1.50$. <br> ACOS H.R. 1000 ohms. 53p. Mono Stethoscope 8 ohms. 53 p . EAR PIECES. Crystal 25p. Magnetic 8 ohms. 13 p . Magnetic 250 ohms .23 p .

## BRITISH MADE STEREO

 MULTIPLEX DECODERBrand New. 7 transistors plus integrated circuit. Fibre glass printed circuit board. Size 21 6t $\times$ in. Pre-aligned. Comd.c. operation. 400 mV output for $100 \mu \mathrm{~V}$ input. Full instructions for any FM Toner. Some technical experience
engential.
$\mathbf{6} \mathbf{5 0}$ DIPOLE LOFT AERIAL $£ 1 \cdot 50$.
CABLE 4p yard.


## AM-FM/VHF TUNING GANG



Super quality small size $1 \frac{1}{1} \times 1 \frac{1}{6}$ $1 \frac{1}{2}$ in., plus spindle $1 \frac{1}{5} \times \frac{1}{5} .385+$ 365 pF with $25+25 \mathrm{pF}$. British made. Geared stow motion drive 6:1. Plastic dust cover. 6BA tapped front fixing. Cast aluminium frame

50p Poat free

E．M．I．WOOFER AND

Compriaing a fine example of a Wooler Comprising a fle example of a wooter Hagnet，44oz，Gauss 13，000 lines． Alaminium Cone centre to improve middle and top response．Also the E．M．I Tweeter 3 in．aquare has a special lightweight paper cone and magnet flax 10，000 lines．With Crossover condenser． Impedance Standard 80 oms $\begin{array}{ll}\text { Maximam power } & 12 \text { watts } \\ \text { Uselul Response } & 35 \text { to } 18,000 \mathrm{cps}\end{array}$ Bass Resonance $\quad 45 \mathrm{cps}$ SUITABLE ENCLOSURE $20 \times 13 \times 9 \mathrm{in}$ ．


BAKER 12in MAJOR £9
$30-14,500$ c．p．s．， 12 in ． tweeter cone together with a BAKER ceramic magnet assembly having a flax density of 14,000 gatss and a total flux of 145，000 Maxwells．Bass resonance 40 c．p．$\%$ ．Rated 20 walts．Scales or

Module kit，30－17，000 c．p．s． with \｛weeter，crolsover
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Robustly constructed to atand up to ong periods of electronic power As used by leading groups． Uselul response $30-13,000$ eps． Bass Remonance 55 cps ．

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For 12in．dia．or $\subset 9$ Pos sparer

## 8in．ELAC HI－FI SPEAKER

Dual cone plasticised roll sur－ round．Large ceramic magnet． $50-16,000$ сps．Bass resonance 55 cps． 8 obm music power
$\{3 \cdot 75$
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MODEL ．＇B＇＂． $16 \times$ For $13 \times 8 \mathrm{in}$ ，or $\times 10 \times 9$ in 8 in ，speaker $\leq 5{ }_{25 \mathrm{p}}^{2}$ MODEL＂C์＂， $16 \times 8 \times 6 \mathrm{jn}$. For $10 \times 6 \mathrm{in}$ ．or $\quad \mathbf{~} \quad \mathbf{4} \begin{aligned} & \text { Posit } \\ & 25 \mathrm{p}\end{aligned}$
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| Key＂T＂for |  | 1 1／in．Round 98p | 2in．Round 12.05 |
| a bove | 5p | 1 in ．Round 98p | 23 in．Round 52.30 |
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| Ein．Round | 83p | 13 Sin ．Round $£ 1.05$ | above ．．．．12p |
| sity．Round | 83p | 1 lin．Round 81.05 | 2 sin ．Round 84.10 |
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E8 ？

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## $\notin 10$ 紅



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## EI5 5



## AUDITORIUM

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€ 14 揩

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| $1 \mu \mathrm{~F}$ | 450 V | 19p | 1,000 ${ }^{\text {F }}$ | 25 V | ${ }^{27} \mathrm{p}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
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| $8 \mu \mathrm{~F}$ | 450 V | 16p | 2,000 F | 50 V | 53p |
| $16 \mu \mathrm{~F}$ | 450 V | 17p | 2,500 ${ }^{\text {F }}$ | 25 V | 45p |
| $25 \mu \mathrm{~F}$ | 25 V | $7 p$ | 2,500 F | 50 V | 60p |
| $25 \mu \mathrm{~F}$ | 50 V | 8p | 3,000 ${ }^{\text {F }}$ | 25 V | 48p |
| $32 \mu \mathrm{~F}$ | 450 V | 24p | 5,000 ${ }^{\text {F }}$ | 25 V | 55p |
| $50 \mu \mathrm{~F}$ | 50 V | 10p | 5,000 ${ }^{\text {F }}$ | 50 V | 98p |
| $100 \mu \mathrm{~F}$ | 25 V | 10p | 8-8 $\mu \mathrm{F}$ | 450 V | 18p |
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| $250 \mu \mathrm{~F}$ | 25 V | 12p | $16-16 \mu \mathrm{~F}$ | 450 V | 27p |
| $250 \mu \mathrm{~F}$ | 50 V | 17p | $16-32 \mu \mathrm{~F}$ | 450 V | ${ }^{63} \mathrm{P}$ |
| $500 \mu \mathrm{~F}$ | 25 V | 18p | $32-32 \mu \mathrm{~F}$ | 450 V | ${ }^{49} \mathrm{P}$ |
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| $3 \cdot 3 \mu \mathrm{~F}$ | 63 V | p | $68 \mu \mathrm{~F}$ | 15V | 6 p |
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| Jack, $2 \frac{1}{2} \mathrm{~mm}$ screened | 10p |
| Jack, $3 \frac{1}{2} \mathrm{~mm}$ unscreened | ${ }^{8} \mathrm{p}$ |
| Jack, $3 \frac{1}{\frac{1}{m m} \text { screened }}$ | 12 p |
| Jack, tin unscreened | 12 p |
| Jack. ${ }^{\text {din sereened }}$ | 20p |
| Jack, stereo, unscreened | 20p |
| Jack, stereo, sereened | 35p |
| Phono, plastic rop | 5 p |
| Phono, plated metal | $12 p$ |
| Phono, fitsed 4 ft lead | ${ }^{8 p}$ |
| Wander, red or black | 3 p |
| Banana 4 mm . red or black | 6p |
| LINE SOCKETS |  |
| Car aerial | 14p |
| Co-axial | 17p |
| D.I.N. 2 pin (speaker) | 15p |
| D.I.N. 3 pin | 16p |
| D.I.N. 5 pin, $180^{\circ}$ | 16p |
| D.I.N. 5 pin, $240^{\circ}$ | $16 p$ |
| Jack, $3 \frac{1}{2} \mathrm{~mm}$ | 15p |
| Jack, in-screened | 49p |
| Jack, stereo, screened | 30p |
| Phono, plated metal | 14 p |


| CAPACITORS |  |  |  | $0.0027 \mu \mathrm{~F}$ | $\begin{aligned} & 500 \mathrm{~V} \\ & 500 \mathrm{~V} \end{aligned}$ | $\mathrm{S} / \mathrm{M}$ | $\begin{array}{r} 15 p \\ 50 \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2.2pF | 500 V | S/M | 71p | $0.0033 \mu \mathrm{~F}$ | 125 V |  | $6 p$ |
| 3.3 pF | 500 V | S/M | 7 fP | $0.0033 \mu \mathrm{~F}$ | 500 V | Poly. | 6 p |
| 5 pF | 500 V | 5/M | $7{ }^{19}$ | $0.0033 \mu \mathrm{~F}$ | 1.000 V | MDC | 6p |
| 10 pF | 125 V | P. 5. | ${ }^{5} \mathrm{P}$ | $0.0036 \mu \mathrm{~F}$ | 500 V | 5/M | 15 p |
| 10pF | 500 V | S/M | $71 p$ | $0.0047 \mu \mathrm{~F}$ | 125 V | P.S. | 9 p |
| 15 pF | 125 V | P.S. | 5 P | $0.0047 \mu \mathrm{~F}$ | 500 V | Poly. | 6p |
| 15pF | 500 V | Cer. | 4 p | $0.0047 \mu \mathrm{~F}$ | 500 V |  | 20p |
| 18 pF | 500 V | S/M | 71 p | $0.0047 \mu \mathrm{~F}$ | 1.000 V | MDC | ${ }^{\text {p }}$ |
| 22pF | 125 V | P. 5. | 5p | $0.005 \mu \mathrm{~F}$ | 100V | Mylar | ${ }_{5} \mathrm{p}$ |
| 22 pF | 500 V | S/M | $7{ }_{7}{ }^{\text {P }}$ P | $0.005 \mu \mathrm{~F}$ | 500 V | Cer. | 5 p |
| 25pF | s00V | S/M | 7 P | $0.0088 \mu \mathrm{~F}$ | 125 V | P.S. | $10 \frac{1 p}{\text { P }}$ |
| 27 pF | soov | Cer | 4p | $0.0068 \mu \mathrm{~F}$ | 500V | S/M | 30p |
| 33pF | 125 V | P.S. | 5p | $0.0068 \mu \mathrm{~F}$ | 500 V | Poly | ${ }^{6 p}$ |
| 33 pF | 500 V | S/M | 71 P | $0.0082 \mu \mathrm{~F}$ | 125 V | P.S. | $10 \frac{1 p}{}$ |
| 39 pF | 500 V | S/M | 71 p | $0.0082 \mu \mathrm{~F}$ | 500 V | $5 / \mathrm{M}$ | 30 p |
| 47 pF | $125 V$ | P.S. | Sp | $0.01 \mu \mathrm{~F}$ | 18 V | Disc | 4p |
| 47pF | 500 V | Cer | 4p | $0.01 \mu \mathrm{~F}$ | 125 V | P. 5 . | $10 \frac{1}{p}$ |
| 50 pF | 500 V | S/M | 71p | $0.01 \mu \mathrm{~F}$ | 160 V | Poly. | 4p |
| 56 pF | 500 V | S/M | $7{ }^{\text {P }}$ P | 0.01 1 F | 250 V | M.F. | 3 p |
| 68 pF | 125 V | P.S. | ${ }_{7}{ }^{\text {P }}$ | $0.01 \mu \mathrm{~F}$ | 400 V | Poly. | 3p |
| 68pF | 500 V | S/M | $7{ }^{7}$ | $0.01 \mu \mathrm{~F}$ | 500 V | Cer. | 5 p |
| 75 pF | 500 V | S/M | ${ }_{7}{ }^{1} \mathrm{p}$ | $0.01 \mu \mathrm{~F}$ | 500 V | S/M | ${ }^{30} \mathrm{p}$ |
| 82 pF | 500 V | S/M | 7 p | $0.01 \mu \mathrm{~F}$ | 600 V | MDC | $7 \mathrm{7p}$ |
| 100 pF | 125 V | P.S. | ${ }_{7} 5$ | $0.01 \mu \mathrm{~F}$ | 1.000 V | MDC | 9 p |
| 100pF | 500 V | S/M | $7{ }^{1+p}$ | $0.015 \mu \mathrm{~F}$ | 160 V | Poly. | 3 p |
| 100 pF | 500 V | Cer | 5p | $0.015 \mu \mathrm{~F}$ | . 400 V | Poly. | 3 p |
| 20 pF | 500 V | S/M | $7{ }^{\text {P }}$ | $0.02 \mu \mathrm{~F}$ | 100V | Mytar | 3 P |
| 50pf | 125 V | P.S. | 5 P | $0.022 \mu \mathrm{~F}$ | 18V | Disc | 5 P |
| 150 pF | 500 V | S/M | $7{ }_{\text {¢ }}{ }^{\text {P }}$ | $0.022 \mu \mathrm{~F}$ | 250 V | M.F. | 3 l |
| 150 pF | 500 V | Cer. | 5p | $0.022 \mu \mathrm{~F}$ | 400V | Poly. | 3 p |
| 1800pF | 500 V | S/M | ${ }^{71 p}$ | $0.022 \mu \mathrm{~F}$ | 600 V | MDC | 7 tp |
| 200 pF | 500 V | S/M | $7{ }^{5} \mathrm{p}$ | $0.022 \mu \mathrm{~F}$ | 1.000 V | MDC | 9 p |
| 220pF | 125 V | P.S. | 5p | $0.033 \mu \mathrm{~F}$ | 250 V | M.F. | P |
| 220 pF | 500 V | Cer | 5 p | $0.033 \mu \mathrm{~F}$ | coov | Poly. | 4 p |
| 250pF | 500 V | S/M | 8p | 0.047 $\mu \mathrm{F}$ | 12 V | Disc | ${ }^{6 p}$ |
| 270 pF | 500 V | Cer | 5p | $0.047 \mu \mathrm{~F}$ | 160 V | Poly. | 3 p |
| 300 pF | 500 V | S/M | ${ }^{8 p}$ | $0.047 \mu \mathrm{~F}$ | 250 V | M.F. | 3 p |
| 330 pF | 125 V | P.S. | 5 p | $0.047 \mu \mathrm{~F}$ | 400V | Poly. | 4 p |
| 330 pF | 500 V | S/M | 8p | $0.047 \mu \mathrm{~F}$ | 600 V | MDC | 8 p |
| 390pF | 500 V | 5/M | ${ }_{5}{ }^{\text {p }}$ | $0.047 \mu \mathrm{~F}$ | 1,000V | MDC | 0 p |
| 470 pF | 125 V | P. 5. | 5p | $0.1 \mu \mathrm{~F}$ | 30 V | Disc | $6 p$ |
| 470 pF | 750 V | Disc | 5 p | $0.1 \mu \mathrm{~F}$ | 250 V | M.F. | 4 p |
| 500 pF | 500 V | S/M | 8 p | $0.1 \mu \mathrm{~F}$ | 400 V | Poly. | 5 p |
| 560 pF | 500 V | 5/M | 8p | $0.1 \mu \mathrm{~F}$ | 600V | MDC | 10 p |
| 680 pF | 125 V | P.S. | ${ }^{6 p}$ | $0.1 \mu \mathrm{~F}$ | 1,000V | MDC | 13 p |
| 680 pF | 500 V | S/M | ${ }^{8 p}$ | $0.15 \mu \mathrm{~F}$ | 250 V | M.F. | 5 p |
| 820pF | 500 V | 5/M | ${ }^{8 p}$ | $0.22 \mu \mathrm{~F}$ | 150 V | Poly. | ${ }^{6 p}$ |
| $0.001 \mu \mathrm{~F}$ | 100 V | Mylar | ${ }^{3 p}$ | $0.22 \mu \mathrm{~F}$ | $250 V$ | M.F. | 5 p |
| $0.001 \mu \mathrm{~F}$ | 125 V | P.S. | 6 p | $0.22 \mu \mathrm{~F}$ | 400 V | Foil | 10 p |
| $0.001 \mu \mathrm{~F}$ | 400 V | Poly | ${ }^{3 p}$ | $0.22 \mu \mathrm{~F}$ | 1.000 V | MDC | 15 p |
| $0.001 \mu \mathrm{~F}$ | 500 V | S/M | 10 p | $0.33 \mu \mathrm{~F}$ | 250 V | M.F. | 8 p |
| $0.001 \mu \mathrm{~F}$ | 500 V | Cer | 5p | $0.47 \mu \mathrm{~F}$ | 250 V | M.F. | ${ }_{5}{ }^{\text {p }}$ |
| $0.001 \mu \mathrm{~F}$ | 1,000V | MDC | 6 p | $0.47 \mu \mathrm{~F}$ | 400 V | Foil | 15 p |
| $0.0015 \mu \mathrm{~F}$ | 400 V | Poly | 3 p | $0.47 \mu \mathrm{~F}$ | 1.000 V | MDC | ${ }^{20} \mathrm{p}$ |
| $0.0015 \mu \mathrm{~F}$ | 500 V | S/M | 10 p | $1.0 \mu \mathrm{~F}$ | 250 V | M.F. | $15 p$ |
| $0.0015 \mu \mathrm{~F}$ | 500 V | Cer. | 5p |  |  |  |  |
| $0.0018 \mu \mathrm{~F}$ | 500 V | S/M | 10p | Note: |  |  |  |
| $0.002 \mu \mathrm{~F}$ | 100 V | Mylar | 3 p | 5/M |  |  |  |
| $0.002 \mu \mathrm{~F}$ | 500 V | Cer. | 5 p | P.S. $=1$ | ystyr | = |  |
| $0.0022 \mu \mathrm{~F}$ | 125 V | P. 5. | ${ }^{6 p}$ | MDC | c. rad | $=3$ |  |
| $0.0022 \mu \mathrm{~F}$ | 500 V | S/M | 10p | M.F. $=$ | fard | n. |  |
| $0.0022 \mu \mathrm{~F}$ | 1,000 V | MDC | $6 p$ | Cer. | mic. |  |  |

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[^0]SATURDAY FROM


Summer was so late arriving this year that it nearly slipped by without our noticing it! Already we are tucking the lawn mower in for its winter hibernation, and once again "the nights are drawing in" HOORAY! -that means we can really get cracking on those electronic projects that have been buzzing around at the back of our minds.

The first step is to get the components lined up. And the best way to do that? Simple-get a copy of the Home Radio Components Catalogue. Then we can quickly and clearly see what we need and how much it costs. We'll know too that either by sending an order or by calling in at the Mitcham shop, we can get just what we want "off the shelf". The catalogue costs 70 p , including post and packing, or 50 p over the counter. Every copy contains 10 vouchers, totalling a value of 50 p when used as instructed-so you can soon get back the cost of the catalogue. Over 8,000 items are listed, over 1,500 of them illustrated, and free supplements are supplied regularly to keep you up to date.

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## PROBLEMS OF PROGRESS

THE pace of technical progress has become so accelerated that many interesting developments and techniques, especially in the field of integrated circuits, hardly see the light of day before their demise is threatened by yet another new development. Innovation leap-frogs innovation, and consequently valuable and thought-provoking devices become redundant long before they have been fully exploited. This poses problems to industrial users of such components, and despite crystal-ball gazing and shrewd planning ahead, many equipment makers have been caught out by this explosive state of the art.

The break-through in small calculator designs based on the single-chip i.c. is an outstanding case at this particular time. Commercial desk calculators costing in the region of $£ 250$ have been replaced overnight by new instruments selling at one third the price. Business people who purchased their calculator last year must be wringing their hands and wishing they had waited until the spring of '72. But it is easy to be wise after the event.

The same problem, although on a smaller scale, affects a magazine such as ours, catering for the private constructor. Does one sit back neglecting contemporary and readily available devices, just waiting for the "ultimate" one to turn up? For example, to have knowledge of impending changes in i.c. design that are likely to revolutionise the commercial calculator in the near future is one thing; it is quite a different thing, though, to predict with any certainty if and when these new l.s.i. chips will be made available to private individuals -and if so at what cost.

In such circumstances the P.E. Digi-Cal desk calculator was commissioned. The abundance of cheap TTL devicess seemed to justify such a project being presented to the private constructor. The total cost of material for this project made it a most favourable proposition in view of the then prevailing cost of a comparable commercial instrument. However, between the completion of this design and the publication of the first article, the big break-through in cheap commercial calculators occurred.

Though "priced out of the market" the Digi-Cal remains a unique and valuable contribution to the record of design information published expressly for the benefit of private constructors and electronics enthusiasts. As an education exercise, the real worth of this project cannot be denied. Certainly it offers the technically inquisitive far more than the mere assembly of any commercial kit can possibly provide.

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## SPECIAL DATA SHEET

## I.C. LEAD-OUT IDENTICHART

We apologise for any slight loss of quality of print in this month's issue. Supplies of the normal paper were not avaifable due to the Docks Dispute.

Our November issue will be published on Friday, October 13

[^1]Mosr of the cheaper intercoms on the market at the present time have a low output which can be very distorted. In order to overcome these difficulties it was decided to build a battery powered intercom using an integrated circuit chosen from the General Electric range of audio amplifiers.

The i.c. chosen was the PA234 which has a quoted output of just under one watt when working from an 18 V supply with a typical distortion of three per cent. A preamplifier is needed to drive this i.c. and a simple, one transistor, common base type is used.

Even though the basic system described here is only a simple master/slave one it is very easy to extend it to master/master or master and two or more slaves. with or without call facilities; the necessary alterations for these additions are given at the end

As no commercial cases could be found which looked right, it was decided to build these as well.

## CIRCUIT OPERATION

The amplifier circuit diagram is shown in Fig. 1. It will be seen that TRI is connected in the common base mode which is relatively rare.

In this configuration the signal is applied to the emitter producing the advantages of low input impedance-of the order of approximately 50 ohms, and a high output impedance, thus making it ideally suited to its purpose.

The microphone/speaker is a 25 ohm loudspeaker and this is easily matched to the preamplifier stage, another advantage being that this configuration prevents inductive pickup from TV sound. Whilst experimenting with the prototype a standard high gain common emitter preamplifier was used and the TV pickup drowned all other signals.

The purpose of Cl is to filter out any TV or other radiated interference; the final version gives

no mains hum even though the cables run alongside mains leads for some distance.
The biasing for TR1 is provided by $\mathrm{R} 1, \mathrm{R} 2$ and C 2 which ensures that the base is near earth potential as far as alternating signals are concerned. R3 serves a dual purpose in that it reduces the supply of 18 V down to 9 V for the preamplifier and in conjunction with C3 decouples the two stages preventing any feedback via the supply.
Potentiometer VRI is a volume control and should be adjusted to give the best output. The signal is then fed via C4 and R4 to the input terminal, pin 3. of the i.c. The ratio of R5 to R6 is determined from the manufacturer's data sheet and controls the correct biasing of the i.c., whilst the gain is determined by the ratio R5 to R4.
The output is stabilised against oscillation by the capacitor C5, and C7 further stabilises the arrangement.
The circuit for the call oscillator is given in Fig. 2 and it will be seen that it is a conventional multivibrator working at about 1 kHz . Other frequencies can be selected by varying R8, R9 and C8, C9.

The power supplies are two PP6 batteries connected in series to give 18 volts.

## MASTER/SLAVE SWITCHING

Fig. 3 shows a simple master/slave set up. S I is a miniature lever key switch having four sets of changeover contacts on each side. This switch is wired so that when the key is in the central position the master loudspeaker is in the listen mode. If now $\mathbf{S} 2$ is operated then power is applied to the amplifier and the slave cat talk to the master.

Should the master wish to call the slave then pushing the key one way the oscillator which provides the calling tone and the amplifier are both powered and the output is fed to the slave speaker.

Pushing the key the other way switches the slave to the listen position thus enabling the master station to speak. This arrangement of switching ensures that privacy is enjoyed by the slave station-the master can not easily listen in, though, if desired, by mounting S 2 at the master end one can control all functions from the one position.

The wiring of the switch is shown in Fig. 7.


Fig. 1. Circuit diagram of the main amplifier


Fig. 3. Diagram showing the switching arrangement for a simple master/s!ave ksystem. Only four wires connect the master to the slave (labelled $A, B, C, D$ ). Switch $S 2$ and loudspeaker LS/MIC2 are in the slave


Fig. 2. Circuit diagram of the call oscillator

## By R. A. PENFOLD



## AMPLIFIER BOARD

\section*{Resistors <br> | R1 | $10 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $1 \mathrm{M} \Omega$ |
| R3 | $10 \mathrm{k} \Omega$ |
| R4 | $100 \mathrm{k} \Omega$ |
| R5 | $680 \mathrm{k} \Omega$ |
| R6 | $100 \mathrm{k} \Omega$ |
| All | $\pm 10 \%, \frac{1}{4} \mathrm{~W}$ carbon |}

Potentiometer
VR1 $10 \mathrm{k} \Omega$ log.
Capacitors
C1 $4,700 \mathrm{pF}$
C2 $6.4 \mu \mathrm{~F} 25 \mathrm{~V}$ elect.
C3 $100 \mu \mathrm{~F} 25 \mathrm{~V}$ elect.
C4 $0.22 \mu \mathrm{~F}$
C5 $0.047 \mu \mathrm{~F}$
C6 $250 \mu \mathrm{~F} 30 \mathrm{~V}$ elect.
C7 $0.1 \mu \mathrm{~F}$

## Semiconductors

TR1 2N3706
IC1 PA234 (General Electric)


Speakers
LS/MIC1, LS/MIC2 22S 3in (2 off)

## Switches

S1 Miniature lever Keyswitch, 4 pole changeover in each position
S2 Push on, release off push button

## COMPONENT ASSEMBLY

The components are mounted on 0 lin pitch Veroboard as shown in Fig. 4 and Fig. 5, and special care is needed to ensure that no solder bridges are formed between adjacent strips. No real trouble should be encountered in the construction providing the diagrams showing the layout are followed carefully and that all breaks are made.

When soldering the i.c. in position do not leave the iron in place for too long. The i.c. needs a heat sink of at least 1 sa in and this can be made from atn offcut of phosphor bronze draught strip though any copper or tin sheet will do. This heat sink, which should be bent so that it does not touch other components, is soldered directly onto the tab attached to the i.c.-take great care not to bend this tab too much or else it will break off and render the i.c. useless.

## CASE CONSTRUCTION

The case for the master station was made from 20 gauge aluminium sheet. Fig. 6 gives the various details of sizes and cuts to be made.

A number of home constructors are put off making their own cases by the thought that it is too difficult and that the finished job will look "home made". Provided that the aluminium is marked out accurately and carefully before any cuts or bends are made no difficulty should be encountered.

When all the various bits have been cut out an Abra-file can be used to cut the hole for the loudspeaker and key switch. The exact size of these will depend on the actual components obtained. The holes for fixing the key switch should be drilled at the same time.

The inside of the master box is seen here with the components removed from their positions. The brackets which hold the potentiometer and the call oscillator board can be seen fixed to the inside


## COMPONENTS . . .



Fig. 5. Layout of the components on the call oscillator board

Fig. 6. Construction of the case for the master. 20 s.w.g aluminium should be used and cut and bent to the . sizes shown



## ASSEMBLY OF THE UNIT

When fitting the switch it is necessary to wire it up first and then fit it in position.

Any offcuts left can be used to make the bracket for VR1 which is glued in position using Evostick or Araldite. Other pieces can be used to make the circuit board support. The terminal block is also glued to the inside of the case on the opposite side to VR1 as shown in Fig. 7.

The case for the slave station can be purchased, adapted from an available tin or constructed from aluminium. It should be covered in the same way as the master station.

The amplifier circuit board should then be

iq. 8. Switching arrangement to give multiple slave operation, though none of the slaves has a call facility. The "select slave" switches are push buttons mounted on the master box
mounted on the bracket shown using insulating washers to ensure that the Veroboard strip is not shorted. The call oscillator can be mounted on its bracket which is then glued to the inside of the case behind the terminal block; to aid adjustment it helps if the preset on this board is easily accessible.

## TESTING THE FINISHED UNIT

Before commencing testing ensure that all components are in the correct position, that their values and polarities are as in the diagram and also that you have made all the breaks. Trace all wires to see that they do go to the right terminals and most important that no solder bridges are present.

When you are completely satisfied connect the external speaker and plug the batteries in. It helps


The internal construction of the slave showing the speaker and the push button in place
if a working transistor radio is placed near the slave station, when the switch is pushed one way you should hear the radio-adjust VRI to give the best results. If testing is carried out with the stations too close together, feedback will occur giving an almighty howl.

Should you have no helper available you can short the two terminals taking power to the slave button at the master terminal block, this will be the same as pressing the "talk" button at the slave. Should there be a lot of distortion occurring when the slave talk button is pressed and the only cure is to reduce the volume you can try decoupling the two power lines going to the slave with a capacitorsay $50 \mu \mathrm{~F}$-connected across the appropriate terminals at the station end. When the lever key is pushed towards "call" the tone should sound in the slave speaker. The volume and quality can be adjusted to some extent by altering the value of the preset on the oscillator board.
continued on page 839


Completed master unit


## GRAVITATIONAL WAVES

Some time ago details of an experıment to detect gravitational waves by radio was described. This experıment, which was originally suggested by Dr John Jelley, made some stir when the results were reported in Nature. There were two series of the experiments but the results were negative. There were certain difficulties mainly due to the high level of noise on the frequency of operation.

The original paper on gravitational waves was put forward by Joseph Weber in 1969. He claimed at that time that he detected these waves as coming from the centre of the Galaxy.
The possibility of these waves was predicted by Einstein in 1916 but nothing much was attempted in the way of investigation. The principle which underlies the theory is that gravitational waves are a form of radiation which affects mass. Small displacements of mass are caused by the radiation setting up a definite vibration. If they, are in line with the radiation the masses do not move with respect to each other.

Professor Weber's experiment was made with two huge aluminium cylinders deep down in the earth. Though Weber is satisfied that his results are positive, and indicate quite definitely that these waves come from the galactic centre, not all the scientific community accepts this.

## SOURCE CONFIRMED

It is, therefore, quite a move forward when independent experimenters confirm the source of the waves. Three physicists from Israel have published some work on their
experiments and it seems to be a big step forward in this field. They are using the earth itself as the detector. Its great mass is much more sensitive than arrays of discs or cylinders.

If these waves are in fact coming from a specific point in the galaxy, then using the earth as a detector, all those points in line with the waves would vibrate the earth. Since the earth revolves on its axis this particular point on the earth would show a peak of activity twice in a sidereal day.

This effect will occur every ilhr 58 min and has been detected by Dr Sadeh of Tel Aviv and Dr A. Ben-Menahem of the Weizmann Institute. They have set up a seismometer in the fairly noise free area on the Gulf of Aqaba.

The team set out to discover if pulsars were the cause of the gravitational waves; pulsars are dense stars and revolve very rapidly. Since it is unlikely that they are completely homogeneous, there will be angular moments which will show as a variation of the gravitational effect. In other words there will be a certain frequency of change which could be detected.

## FIRST EVIDENCE

The first evidence that this was right thinking came in 1971. The Israeli physicists found that there was a peak in the recording which corresponded to a periodicity of $0.53-0.60$ seconds. Moreover, as each day passed the event came a little earlier in time. This was a clear indication that the origin of this event was outside the earth.
The twice a day Iresponse confirmed that the position of the source was at a celestial longitude of Right Ascension 11 hr 33 min . This position corresponded to the pulsar known as CP 1133.

As a check on this the team have examined the radio period of the pulsar and find that their events are in agreement. According to the present model of the pulsar, the radio pulses occur once with each revolution of the star. Gravity waves however would be emitted twice every revolution. Measurements confirmed that the seismometer events did occur at twice the rate of the radio pulses from CP 1133. The actual period of the gravitational waves is 0.594 seconds.

It would seem that there is a chance for confirmation of Weber's untiring efforts in this field. Certainly the benefits to astrophysics will be very great.

## EARLY LINK-UP

It would seem possible that the planned space link-up exercises between America and Russia will be undertaken earlier than was previously expected. While the main
exercise remains set for 1975 between a Soyus and Apollo vehicle, it is possible that a Salyut orb:tal station might be used.

The problems that have to be solved are many. Not the least of these is that of communications. The languages are different, the mode of procedures between the crews and ground will have to be rationalised to name two aspects.

There is also the problem of the airlock system. The A pollo atmosphere is nearly pure oxygen, at a pressure of 5 lbs per square inch, while the Russians use a mixture of nitrogen and oxygen at the normal atmospheric pressure of sea level.

## LUNA-20

Some new evidence about the lunar rock formation has been brought back by Luna-20, the automatic probe sent by the USSR to collect samples from the Moon.

The samples came from the highland area between the Mare Crisium and the Mare Fecunditatis. The soil in this region is much lighter and contains fragments of minerals and crystalline rocks. It is rich in aluminium, calcium and magnesium oxide and the surfaces are well preserved. The iron and titanium content is less than that brought back from the actual maria by Litha-/6.

Altogether some 70 elements have been detected in the lunar soil by Soviet scientists. They are particularly interested in the nature of the lunar iron.

It seems that the lunar iron is even more rust proof than the meteoritic iron, and attempts are being made to simulate lunar conditions on earth with the possible benefits to industry. Already some success has resulted from the sublimation of iron from basalts which have been heated to a high temperature in a vacuum.

## MYSTERY MOON

There are some rather puzzling features about Titan, Saturn's largest moon. Theoretically the gases should have boiled off, since this body has a low escape velocity.

The present low density atmosphere of molecular hydrogen and methane, from which the gases must be escaping quite rapidly into space, must have come into being long after the solar system was formed. This means that there must be a constant replenishing probably by evaporation and gassing from the solid surface.
Measurements made by L. Tafton of the McDonald Observatory suggest that there is some other major constituent in the atmosphere. It also seems probable that methane is present in the solid as well as in the gaseous state.


A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.
This is YOUR page and any idea published will be awarded payment according to its merits.

## TTL EQUIVALENCE GATE

|  | A | B | output |
| :---: | :---: | :---: | :---: |
| same | 0 | 0 | 0 |
| different | 1 | 0 | 1 |
| different | 0 | 1 | 0 |
| same | 1 | 1 | 0 |



Fig. 1. TTL equivalence gate and truth lable

HAVE built and tested an equivalence gate for use in TTL circuitry, see Fig. I. It uses one inexpensive integrated circuit package, type $S N 7400$, a quadruple 2 -input NAND gate.

If the inputs $A, B$ are the same, the output goes low ("0" in positive logic). If the inputs are different the output goes high ("1" in positive logic), see truth table in Fig. 1.

## S. L. Thompson, Rainham.

ANY audio circuit which produces a simulation of an instrument has not only to copy the frequency and waveform produced by the instrument but also has to emulate the general shape of the envelope, because the sound produced from any instrument, when it is played, does not immediately attain full volume but is a gradual build up. It also dies away slowly afterwards. It is in this respect that most simple electronic devices are inaccurate, as they produce an envelope with virtually no rise or decay time.

The circuit shown in Fig. 1 will produce a note with a rise and decay time variable between two limits. With the values given the note produced is approximately 830 Hz ,
The first stage TR1 and TR2 are in the normal multivibrator configuration and the waveform which should be obtained from point A is shown in Fig. 2. Transistor TR3 is an overdriven amplifier to produce a good square wave and Fig. 3 shows the waveform that should be obtained at the output, point B. Transistor TR4 acts as an attenuator and is controlled by C3, VR1 and VR2 connected to its base.

The output from the circuit is taken from TR4 collector ( $R_{\mathrm{L}}$ ), and if a resistive load is used its value should be $5.6 \mathrm{k} \Omega$, but other loads can be used to give different tones. An inductive load gives a sawtooth waveform and has a peculiar metallic tone. Other loads, such as capacitive, could be tried.

The envelope produced is shown in Fig. 4; $t_{\mathrm{r}}$ is the rise time, which will occur if $S 1$ is pressed at time 0 . Potentiometer VR2 gives a rise time variable between $0-50$ milliseconds. The decay time $t_{\mathrm{d}}$ is controlled by VR1 and starts at point (a) in Fig. 4 and can be varied between $0-200$ milliseconds.

The output should be attenuated before (if desired) it is amplified as it is in the region of 7 V .

Switch SI should be of the push-on, release-off type with position 1 normally closed and position 2 normally open. The switch should be the break-before-make type because if the rise and decay time were both set at 0 , the battery could be shorted out.
N. R. Adams,

Colehill,
Dorset.

Fig. 2


Fig. 1. Circuit of variable attack and decay


Fig. 3


Fig. 4



THE circuit shown above is a 4 -digit combination lock, against which the chances of finding the correct sequence of operation is 1 in 19,958,400, which is about 0.00000005 , making it quite hard to pick!

Fundamentally, it consists of a four stage binary computer, the fifth stage being the alarm flip-flop which is only activated when the wrong digits are dialled. Each flip-flop, when switched, provides the power supply for the proceeding flip-flop and so on up the chain until the relay is activated, and thus driving the lock mechanism.

The flip-flops of the computer are simple bistables, i.e. it has two stable states. When one transistor is conducting the other one is not. The non-conducting transistor is made to conduct by applying a pulse to the base of conducting transistor.

In the combination lock this is accomplished by shorting the base of the first transistor of each stage to the positive line. When the power supply to a flip-flop is switched on, one transistor will always conduct first. This must be the first in each stage (due to inequalities in the circuit).

Operation of the combination lock is quite simple. The first number is dialled in this case number 6 , and the pushbutton switch S2 is pressed momentarily switching on bistable $A$. The second number is dialled and the procedure is repeated. This procedure is repeated with the last two numbers till the lock is operated.

Resetting of the bistable is accomplished by disconnecting the power supply and then reconnecting it.
G. E. Kaparnick, Swindon, Wilts.

[^2]

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| FR65 | Dual Cone | 680 | High Frequency Units |  |  |
| FR8 | DualCone | 8.90 | CT5 | Cone | 120 |
| Crossover Networks |  | CT10/8 | Horn | 240 |  |
| CN23 | 2 way | 2 way | 110 | CT10/16 | Horn |
| CN28 | 2 way | 110 | HT15 | Horn | 240 |
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# (1) The 741 as an ASTABLE 

THere are literally thousands of different integrated circuits on the market, and it would iake every page of every issue of the magazine to make any impression in reviewing the complete field. Integrated circuits can be divided into two very broad categories, those intended for general applications and those designed with a specific function in mind.

The state of the art is such that many integrated circuits are so complex as to prohibit experimentation. Those for TV signal processing, for example, cannot adequately be used unless we are prepared to rewire a TV! Where appropriate such i.c.s usually appear in constructional articles.

In general research and development a selection of TTL and a few linear circuits allows much "routine" work to be achieved quickly. Manufacturers short form catalogues show where to go for "state-of-theart" circuits if they become necessary.

Rather than extract up to the minute information from data sheets and application notes, we hope to show how the "standard" integrated circuits can be used in many cases. We will place the emphasis on designing circuits using integrated circuits and not necessarily on reviewing a different one each month. We feel that this is more useful for the reader who wishes to experiment because he will often be able to use the same integrated circuit more than once.

Readers requiring further details of individual i.c.s and their performance, are recommended to consult manufacturers' data sheets.

## OPERATIONAL AMPLIFIERS

The first generation of linear integrated circuits were the $\mu \mathrm{A} 702$ and 709 operational amplifiers and the 710
comparator. Both the 702 and 709 have their own disadvantages. The 702 has the limited input voltage range of +1.5 V to -6 V and can "latch up" if this is exceeded. It can however be used for many feedback applications from d.c. to 30 MHz .
The 709 is an improvement on the 702's performance by providing a higher input impedance, a large input common mode range of $\pm 10 \mathrm{~V}$ and a high output swing. It has its own characteristics (cross-over distortion, for example) and is not short-circuit proof. It can be used for high performance applications from d.c. to 5 MHz or so.

Both the 702 and 709 require several external components to provide frequency compensation so that they are stable under closed loop conditions, and do not become unstable when overall feedback is applied. This compensation can be chosen to obtain the best bandwidth for a given closed loop gain.
The first of the "second generation" of integrated circuits was the $\mu \mathrm{A} 741$ which overcomes many of the disadvantages of the 702 and 709 series. It is very suitable for general work where a wide bandwidth and a fast slew rate are not required.

This month we introduce the 741C amplifier and will use it again later in the series. Although it is convenient for demonstration purposes its bandwidth restricts its use and the 702 or 709 circuits can be used, within their limitations, when we wish to work outside the frequency limit of the 741.
The 702, 709, 741 integrated circuits are available, under slightly different numbers, from many manufacturers. (Some guidance on these numbers can be seen in the table on the P.E. I.C. Identichart in this issue.) -

## THE 741C OPERATIONAL AMPLIFIER

The 741 C is a high performance operational amplifier intended for a wide range of analogue applications from d.c. to 1 MHz . Its high common mode range and absence of "latch up" make it ideal for a voltage follower.

It requires no external components for frequency compensation and is stable under closed loop conditions including unity gain and integrator applications. It is protected against an indefinite short circuit to ground or to either supply rail. The input offset voltage is typically 2 mV and this maty be nulled by a simple circuit.

Its connections are shown in the diagrams for the various package outlines. Typical parameters are shown in the Table 1.1. Versions without the "C" suffix have slightly improved parameters, such as working temperature range.

## LOW FREQUENCY MULTIVIBRATOR

To start the series, let us look at the 741 wired in an astable multivibrator configuration. Although not necessarily the best application for the 741, it does illustrate how very simple the external circuitry of an i.c. can be to provide a useful application.

Fig. 1.1 shows a basic astable multivibrator suitable for generating a fixed frequency square wave for test purposes. It has a negative feedback loop from the output to the inverting input via R3 and a positive feedback loop from the output to the non-inverting input via R1 and R2. The circuit changes state regeneratively each time the two inputs are at the same potential.

If we assume that the circuit is operating and that the output has just switched to $+V_{0}$ (the positive saturated output level) then the voltage $V_{1}$ at the non-inverting input (marked + ) will depend on R1 and R2. The voltage $V_{2}$ at the inverting input (marked - ) will be rising exponentally towards $+V_{0}$ from an initial negative value.

$$
\text { Let constant } \begin{aligned}
\mathrm{k}=\frac{R_{1}}{R_{1}+R_{2}} \quad \text { Initially } V_{1} & =+k V_{0} \\
V_{2} & =-k V_{0}
\end{aligned}
$$

Wher, $V_{2}$ reaches $V_{1}$ the amplifier will change state with the output switching to $-V_{0}, V_{1}$ switching to $-k V_{0}$, while $V_{2}$ now falls exponentally towards $-V_{0}$ from its initial value of $+k V_{0}$. When $V_{2}$ reaches $V_{1}$ (which is now $-\mathrm{k} V_{0}$ ) the circuit changes state again and the cycle repeats.

Table 1.1: Typical Characteristics of the 741C
Large signal open loop gain 100 dB
Input resistance
Differential input offset voltage Differential input offset current Input voltage range
Maximum output voltage swing
Common mode rejection ratio Input bias current
Output resistance
Supply voltages
Slew rate at unity gain

1 Megohm
2 mV
30 nA
$\pm 13 \mathrm{~V}$
$\pm 13 \mathrm{~V}$
90 dB 200 nA
75 ohms
$+15 \mathrm{~V}$
$0.5 \mathrm{~V} / \mu \mathrm{s}$

## ANALOGUE COMPUTING Computing

 system where continuous signals represent mechanical (or other) parameters.COMMON MODE REJECTION RATIO
A common mode input is one applied to both inputs of an operational amplifier. In theory this should produce no output, since only the difference between the inputs is amplified. The CMRR is a measure of this rejection to a common input signal, and should be large.
COMPARATOR An amplifier designed for comparing two inputs and giving a 0 or 1 output. Usually the output is compatible with logic levels.
DIFFERENTIAL INPUT An input applied between two terminals, neither of which is at earth potential.
differential input rating the maximum differential input which may be applied between the two terminals.


Fig. 1.1. The basic astable multivibrator using a 741

DIGITAL COMPUTING Computing system where two valued signals represent mechanical (or other) parameters.
FREQUENCY COMPENSATION The compensation required in feedback amplifiers to ensure stability and prevent unwanted oscillations.
INPUT COMMON MODE RANGE The maximum input which can be applied to either input without causing damage or abnormal operation.
INPUT OFFSET CURRENT The difference in input bias current requirements for the two inputs.
INPUT OFFSET VOLTAGE That voltage which must be applied between the input terminals to obtain a zero output voltage.
INTEGRATED CIRCUIT An electronic device containing several elements, active or passive, which perform all or part of a circuit function.


Basic connections (looking at top) of inputs and output for the 741. For details of other connections, see the I.C. Lead-out Identichart


Fig. 1.2. A practical circuit for 1 kHz operation

LATCH UP The switching of an electronic circuit to an unintended mode by improper voltage application.
OFFiSET ERRORS Errors produced by input offset voltage and input offset current.
OPEN LOOP GAIN The gain of an amplifier before negative feedback is applied.
OPERATIONAL AMPLIFIER Originally a direct coupled high performance amplifier used in analogue computing to perform mathematical operations (summation, scaling, subtraction, integration, etc.). Today a low frequency packaged amplifier (about d.c. to 1 MHz )
SLEW RATE The maximum rate of change of the output voltage for large signals. Usually in $\mathrm{V} / \mu \mathrm{s}$.

## PERIOD OF OSCILLATION

If we assume that the inputs of the amplifier draw no current and that the positive and negative saturation levels are equal, then the period of oscillation is given by:

$$
\tau=2 C_{1} R_{3} \log _{e} \frac{(1+k)}{(1-k)}
$$

For $\mathrm{R}_{1}$ equal to $\mathrm{R}_{2}$, this becomes $:=2 \cdot 2 C_{1} R_{3}$ secs.
Although there is no need to do so, it is convenient to make $\mathrm{R}_{3}$ equal to the parallel combination of RI and R2 to minimise offiset errors. The maximum allowable value for $k$ depends on the differential input rating of the amplifier. For the 741, values for $k$ of up to 0.9 are suitable. For a 709 with $\pm 12 \mathrm{~V}$ supplies a k of 0.2 would be used.

## PRACTICAL CIRCUIT

A practical circuit for a nominal 1 kHz operation is shown in Fig. 1.2. The Zener diodes clip the output level at $\pm 6 \mathrm{~V}$ to aid slew rate slightly. The measured frequency was 844 Hz . This is 13 per cent lower than the nominal value, probably due to the 20 per cent tolerance of the capacitor.

Rise and fall tinles were $10 \mu \mathrm{~s}$, equivalent to a bandwidth of 35 kHz . Although this is noticeable for a 1 kHz square wave, it is still good enough for simple audio testing to illustrate the effects of tone controls, instability and ringing, or filters. For higher frequencies a 709 can be used with a value of 0.2 for $k$.

At frequencies lower than 1 kHz , or if slightly longer rise and fall times can be tolerated, the Zener diodes may be removed and R4 shorted out, so that the circuit relies on the symmetrical limiting characteristics of the amplifier.

## Part 2 Next month: Introduction to basic logic gates

## A novel entertainment for parties and fêtes

NOUGHTS and crosses is a simple but fascinating game whose rules can be understood even by young children. Two players alternately place noughts and crosses in a three by three grid in an attempt to obtain a complete line of their own symbol either diagonally, horizontally or vertically.

With this machine the player is given the advantage of starting first and makes his play by using a pen to "write" a nought on to a colourful illuminated screen. The computer instantly replies by displaying a cross. Two anti-cheat devices are included to foil baffled players.

Most of the 945 different games that can be played against the computer will be drawn or lost, but in two per cent of the games the player can get three noughts in a line. Finding these winning

games is not easy and presents an interesting prob lem in logic for the player. Schoolchildren will play for hours trying to find one or more ways of beating the computer.

## ANALYSIS OF THE GAME

The object of the game is to obtain a row (horizontally, vertically or diagonally) of three noughts, while preventing the computer from obtaining a row of three crosses.

In the usual game using pencil and paper the first player has a choice of nine places, his opponent then has a choice of eight places, and so on. Thus there are $9 \times 8 \times 7 \times 6 \times 5 \times 4 \times 3 \times 2 \times 1=362,880$ ways of filling the places. Though not all of these represent interesting games, the number is too large for a simple computer and so must be reduced.

This is done by allowing the player any choice of position, but limiting the machine to just one reply The total is then $9 \times 1 \times 7 \times 1 \times 5 \times 1 \times 3 \times 1=945$ possible games, a number which allows simplified circuitry but still provides variety of play.

For further analysis of the game the squares are numbered 1 to 9 as in Fig. 1.

Square 5 is the most important because it controls four of the eight possible rows of three. Squares 1 , 3, 7 and 9 each control three rows. Least important are squares $2,4,6$ and 8 because each can only block two rows.

## TACTICS OF THE MACHINE

The first basic principle of play is that if the player's first move is in one of squares $1,2,3,4,6$, 7,8 or 9 , then the computer's best reply is square 5 If the player's first move is square 5 , then the computer's reply should be either $1,3,7$ or 9 , but in practice it is made to be always square 1

The player's first move determines the computer's reply for each successive move in that game. These moves and the corresponding replies are shown in Fig. 2(a) to (i).

Consider Fig. 2(a). The player has started with square 5 and the computer has replied with square 1 .


The completed prototype showing the general appearance of the unit with the "how to play" notice in position

The double-headed arrows indicate reversible positions; for example if the player plays square 2 the computer answers with square 8 (to block the row), while if the player plays square 8 , the computer replies with square 2 (to block the row).

This illustrates the second design principle: when the player fills in a second " 0 " in any line, the computer should block the line with an " X ".
It will be seen in Fig.2(a) that one of the arrows is single-headed indicating an irreversible play. If the player plays square 9 the reply is square 7 , whereas if square 7 is played the reply is not square 9 , but square 3 .
This illustrates the third principle: if the player is just about to start a line of "0"s, the computer should reply with a second " X " in a line so as to force the player on to the defensive.
If these three principles are followed in the usual pencil and paper game then the player cannot lose.
Possible moves and replies for the eight other starting positions can be followed by reference to Figs. 2(b) to (i).

## INPUT UNIT

The player makes each move by touching the pen to one of nine lin diameter brass washers (corresponding to the nine squares). Each washer is connected to the gates of two thyristors (SCRs).

When a washer is touched, one of these thyristors switches on a lamp indicating the player's " 0 ", and the other thyristor, by means of the memory store, switches on the computer's reply " X ".

Once on, the thyristors will stay on until the power supply is interrupted. Thus the use of thyristors has the important advantage that once a move is made, it cannot be changed either accidentally or deliberately. If toggle switches were used instead, players could cheat by flicking the switches to find the best move at each stage.

Fig. 3 shows the circuit diagram for these thyristors and Fig. 4 shows how they may be mounted on a piece of Veroboard. Eighteen 50 p.i.v. 1A SCRs were used in the prototype, but thyristors with other ratings may be used providing they are all of the
same type to ensure that the lamps light equaliy brightly.

Also needed for the input unit is the pen and a resistor which may be fitted inside the pen. The resistor should be the highest value that will give reliable triggering of the thyristors. A 150 ohm $\frac{1}{4} \mathrm{~W}$ resistor was suitable in the prototype.
The pen is a plastic ballpoint of the kind that unscrews in the middle and has a retracting button at the top end. With the ink tube and button removed, flexible cable may be led in at the top hole, knotted, and soldered to the resistor which in turn is soldered to a thick wire which is then soldered or glued with "Araldite" epoxy resin, so as to pr-trude through the lower hole as in Fig. 5.


Fig. 3. Circuit diagram of the input unit. The thyristors can be any $50 \mathrm{~V}, 1 \mathrm{~A}$ type

## COMPOUENTS

Resistor
R1 $150 \Omega \quad \frac{1}{4} \mathrm{~W}$ carbon (see text)
Thyristors
CSR1-CSR18 Any 50V 1A type (18 off)

## Switches

S1 10 pole 10 way wafer switch (1 wafer must be a break-before-make type)
S2 Push-to-break push button switch

## Lamps and holders

LP1-LP18 M.E.S. lamps and holders (for voltage ratings see text) (18 off)
Miscellaneous
in dia. brass washers ( 9 off), ballpoint pen, large knob, mirror clips ( 6 off), two pieces of glass or Perspex, plywood or hardboard, 2 in $\times 1$ in softwood, red and green cellophane or cinemoid, tracing paper, Araldite (for details of these materials see text), 5 in $\times 1 \frac{1}{4}$ in 0.15 in matrix Veroboard

Fig. 4. Layout of the thyristors on the Veroboard panel


Fig. 5. Internal construction of the "pen"




cathode of


## MEMORY SWITCH

The memory switch consists of a 10 pole 10 way wafer switch. One way is used as an "off" position and the other nine positions of the switch are used to set the initial play number (i.e. select the required programme).

Nine of the wafers are used to route current to the correct lamps and the tenth wafer forms another anti-cheat device. This tenth wafer must be "break-before-make" type. This ensures that if a player tries to cheat by turning to a different programme part way through a game, the power supply to the thyristors is cut off momentarily and the display unit is wiped clean.

Fig. 6 shows the ten wafers. Wafers Sla to Sli must be wired together by reference to the numbers shown round each wafer in Fig. 6. All the contacts marked " 1 " must be wired together (for connection later to the " $X$ " lamp in square 1 of the display unit). Similarly all the contacts marked " 2 " are wired together (for connection later to the " $X$ " lamp in square 2 of the display unit), and so on. Great care should be taken at this stage to ensure correct wiring.

All except one of the contacts of wafer SIj are wired together, as shown in Fig. 6.

## OUTPUT DISPLAY UNIT

This consists basically of 18 m.e.s. lamps in lampholders under a translucent screen. Nine lamps (green) are used to indicate the " 0 "'s and the other nine (red) to indicate the "X"s using filters as described later.



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| No. | (Worts) | $160 z$ |  |  | 1 ¢ D |
| 07 | 20 | 111 | $7.0 \times 6.0 \times 6.5$ |  | 1.6130 |
| 100 | 60 | 38 | $8.9 \times 8.0 \times 7.7$ |  | 2.3936 |
| 61 | 100 | 512 | $10.2 \times 8.9 \times 8.3$ |  | $2.62 \quad 52$ |
| 30 | 200 | 98 | $12.0 \times 10.3 \times 10.0$ |  | 4.3952 |
| 62 | 250 | 124 | $9.5 \times 12.7 \times 11.4$ |  | $5.80 \quad 67$ |
| 55 | 350 | 150 | $14.0 \times 10.8 \times 12.4$ |  | 7.7782 |
| 63 | 500 | 270 | $17.1 \times 11.4 \times 15.9$ |  | 11.20 * |
| 92 | 1000 | 400 | $17.8 \times 17.1 \times 21.6$ |  | 20.63 |
| 128 | 2000 | 630 | $24.1 \times 21.6 \times 15.2$ |  | 34.10 |
| Ref. | VA | AUTO Weight | SERIES (NOT Size cm. | ISOLATED) Auto Tops | $p$ \& $P$ |
| No. | (Wotts) | ) 16 oz |  |  |  |
| 113 | 20 | 11 | $7.3 \times 4.3<4.4$ | 0-115-210-240 | 0.8522 |
| 64 | 75 | 114 | $7.0 \times 6.4 \times 6.0$ | 0-115-210-240 | 1.6630 |
| 4 | 150 | 30 | $8.9 \times 6.4 \times 7.6$ | 0.115-200-220-240 | 2.0036 |
| 66 | 300 | 60 | $10.2 \times 10.2 \times 9.5$ |  | $3 \cdot 8952$ |
| 67 | 500 | 128 | $14.0 \times 10.2 \times 11.4$ |  | 5.7867 |
| 84 | 1000 | 160 | $11.4 \times 14.0 \times 14.0$ |  | 10.4982 |
| 93 | 1500 | 289 | $13.5 \times 14.9 \times 16.5$ |  | 15.20 * |
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 $\begin{array}{llll}11 & 0.5 & 0.25 & 12 \\ 3 & 1.0 & 0.5 & 1\end{array}$ $\begin{array}{llllllllll}131.0 & 0.25 & 1 & 7.6 \times 5.7 \times 4.4 & 0.12 V & \text { at } 0.25 \mathrm{~A} \times 2 \\ 11 & 1.3 & 5.1 \times & 5.1 & 0.12 \mathrm{~V} & 0.25 \mathrm{~A}\end{array}$

[^4] $\begin{array}{r}\text { \& } \\ 5 \\ 2 \\ 22 \\ \hline 22\end{array}$

$$
n
$$



Fig. 7. General layout and dimensions of the complete unit (above)

Fig. 8. Construction of the cabinet and display unit (right)

Fig. 9. Arrangement of the " 0 ' and " $X$ " filters under the tracing paper (bottom right)

## CONSTRUCTION OF CABINET AND DISPLAY UNIT

Fig. 7 shows the general layout. Fig. 8 shows the fascia and glass panels removed.

The dimensions are not critical except that the cabinet must be deep enough (at least $2 \frac{1}{4}$ in) to take the wafer switch with all its wiring.
The lamp compartments should be as small as possible to give a bright display while ensuring that there is enough room to unscrew and replace lamps.
If a compartment is needed for HP2 cells (see "Power Supply" section below) then a maximum size of 11 in $\times 15$ in overall may be used. If an external power supply is to be used then 8 in $\times 15$ in will give a better, brighter display.

The cabinet may be made by glueing and pinning four pieces of 2 in $\times 1$ in softwood to each other and to a rectangular ( 11 in $\times 15 \mathrm{in}$ ) hardboard base, see Fig 8.

Four slotted partitions for the lamp compartments can be cut as shown, the m.e.s. lampholders screwed in position and the partitions slotted together and held in position by a 9 in crosspiece.

A metal plate, drilled to take the fixing bolt of the wafer switch can be screwed in position as shown. A short piece of softwood may be necessary to steady the end of the switch shaft if a battery compartment is required. Terminals or a power lead may be fitted for connection to an external power supply.

## NOUGHTS AND CROSSES

The nine " 0 " lamps are each wrapped in one or more layers of green cellophane and the nine " X "


lamps are similarly wrapped in red cellophane or cinemoid. (Cellophane is obtainable from artists' supply shops and cinemoid, used in stage lighting, is obtainable from large electrical lighting shops.)
The lamp compartments are covered by two sheets of glass or Perspex which sandwich a sheet of tracing paper between them. Also sandwiched (under the tracing paper), are nine large red " 0 "s and nine large green " $X$ "s cut out of cellophane, one of each over each lamp compartment (see Fig. 9). Thus when a green " 0 " lamp lights, it illuminates that square with green light except where the red " 0 " shape absorbs the green light and so appears as a black " 0 ". Similarly when a red " $X$ " lamp lights, a red square with a black "X" appears.

The cellophane or cinemoid should be chosen so that when a red and a green piece are overlapped and held to the light, it should appear as dark as possible without the individual colours being too dense for the lamps. In the prototype it was found that six layers of pale green cellophane and one layer of primary-red cinemoid were most suitable.
The brightness of the display is improved by fitting reflectors of aluminium kitchen foil behind the lamps, taking care not to cause a short circuit between the lampholder terminals.

The two sheets of glass may be held in place by mirror clips.

## FASCIA PANEL

The fascia is cut from $\frac{1}{4}$ in or $\frac{1}{8}$ in plywood or hardboard. Nine holes are drilled under the positions of the nine brass 1 in washers (see Fig. 7). Leads are soldered to the washers, passed through the holes and the washers are glued in position using "Araldite" epoxy resin. They should be numbered 1 to 9 as in Fig. 1.

The "push-to-break" reset switch is fixed to the fascia. A hole is also drilled for the lead to the pen. The wood may be finished with varnish or stain. A carrying handle, and a pen clip may be added.


Fig. 10. Part of the complete circuit to show the method of interwiring the display, memory and input units

A large 12 -faced plastic knob is needed for the wafer switch. The knob should be marked from 1 to 9 plus an "off" position corresponding to the switch positions shown in Fig. 6. A small metal pointer may be fixed to the fascia to facilitate reading the knob.

## INTERWIRING OF THE UNITS

Fig. 10 shows part of the complete circuit.
One side of all the lamps is connected to the positive terminal using thick wire and a circuit similar to a ring main to keep the resistance to a minimum.
The pen lead is also connected to the positive terminal.

The negative terminal is connected through the reset switch to all the contacts on the tenth wafer, using thick wire.
The wiper of the tenth wafer is connected to the common cathode lines on the Veroboard (Fig. 4)
again using thick wire.

The lead from washer 1 is connected to the gates of thyristors CSR1 and CSR2 (Fig. 4 and 10) and similarly with the leads from the other eight washers. Heat shunts should be used to protect the thyristors while soldering.

The anode of thyristor CSR2 is connected to the wiper of wafer Sla (see Figs. 4 and 10). The anodes of the eight other even numbered thyristors are similarly connected. If possible colour coded wire shoul be used here.

A lead from all the switch contacts marked 1 in Fig. 8 is connected to lamp 1 " X " (i.e. the red lamp in square 1). Similarly, all the contacts marked 2 are wired to lamp 2 " $X$ " and so on.

Finally each of the anodes of the " 0 " thyristors is wired to the corresponding green " 0 " lamp (Fig. 4).

## POWER SUPPLY

This must be capable of lighting up to nine lamps in parallel, with least reduction in brightness.

Lead-acid accumulators or a tapping off a car battery or Nife alkaline cells give a bright display
and can be re-charged.

Table 1: POWER SUPPLIES AND LAMPS POWER SUPPLY LAMPS

1 lead-acid accumulator ( 2.2 V ) $1.25 \mathrm{~V}, 0.25 \mathrm{~A}$
2 lead-acid accumulators ( 4.4 V ) $2.5 \mathrm{~V}, 0.3 \mathrm{~A}$ or 3.5 V , 0.3 A

3 lead-acid accumulators ( 6.6 V ) $5.5 \mathrm{~V}, 0.3 \mathrm{~A}$ or 6 V , 0.3 A

2 NiFe cells (2.4V) $\quad 1.25 \mathrm{~V}, 0.25 \mathrm{~A}$
4 HP2 dry cells ( 2 pairs in $\quad 1.25 \mathrm{~V}, 0.25 \mathrm{~A}$
parallel) (3V)
Mains driven d.c. power pack bulb to suit voltage (allow 1 V drop across thyristors)

Table 2: "TRUTH TABLE" FOR CHECKING COMPUTER'S REPLIES

## SWITCH WASHER TOUCHED WITH POSITION

I.C. INTERCOM
continued from page 823


Fig. 9. A call oscillator identical to the one in the master can be used to give the slave a call facility. The switch is a miniature lever switch as used in the master. The letters to the right show how the modified slave is connected to the master

## MODIFICATIONS TO THE BASIC SYSTEM

The basic system can easily be extended by connecting slaves in parallel. Fig. 8 shows a typical set up with three slaves none of which have call facilities to the master; in this case it is the master station which has to initiate any communication by pressing the relevant button. If it is wished to give the slave a call facility then Fig. 9 shows the arrangement and this can easily be wired in if the master/slave set up is sufficient.

If any of the configurations using two or more slaves are built then the cases may have to be redesigned. For example, if it is decided to build the master/two slaves with call facilities then the slaves will have to have a case large enough to accommodate the speaker, the call oscillator and the lever key whilst the master will need to be enlarged so that the multi-button switch and the indicator lights can be fitted.

Depending on what loudspeaker is finally used will dictate the method of fixing; the easiest is to glue them to the case making the hole a dittle smaller than the speaker.

The completed intercom



To achieve a logical building sequence the order of construction presented will be the power supply unit, pre-amplifier and tremolo unit, main amplifier and pitch printed circuit boards. The mounting of the keyboard and other assembly details will round off the series.

To assist in lining up the p.c.b.s the power unit and amplifier are used as test gear, hence the order of construction.

## POWER SUPPLY CIRCUIT

The circuit of the power supply is shown in Fig. 2.1 and is based on well established design principles. A transformer using the 20 volt secondary tap produces an open circuit voltage of approximately 30 volts across a large reservoir capacitor, and this reduces to about 26 volts at 2 amps . Two 6.8 volt Zener diodes connected in series act as the reference voltage with which a portion of the output voltage at the emitter of a ZT3055 is compared. Any error is amplified and fed to the output transistor, thus regulating the output voltage. Capacitor C4 produces a slow rise in output voltage at switch on, whilst C3 and R3 give high negative feedback of any ripple voltage which might be developed at at the output.
The output voltage range of the supply is $16-21$ volts, controlled by VR1, and the maximum output
current is 2 amps . Worst case dissipation in the output transistor TR3 is 18 W with 700 mW in the driver transistor TR2. Two Zener diodes are used rather than one high voltage type in order to reduce the power requirement in one package. TRI and each Zener diode have maximum power dissipation levels of approximately 75 mW .

The earth connector allows the earth to the piano circuits to be disconnected in the event of any instabilities which might occur when the piano is connected to a self earthed external amplifier.

## P.S.U. ASSEMBLY

The small components for the power supply unit are mounted on a simple printed circuit an etching and drilling pattern for which is given in Fig. 2.2.

With the p.c.b. completed and terminal pins inserted first the pre-set potentiometer is mounted, then the resistors, diodes, transistors, capacitors and finally the a.c. input terminal strip as in Fig. 2.3.
The two regulator transistors are mounted on a heatsink and the assembly completed using the transformer for mounting the board. The large capacitors, heatsink and transformer are mounted on a baseboard as seen in the photograph.
After assembly the power supply can be checked for correct operation by measurement of voltage at


Fig. 2.1. Circuit diagram of power supply unit
a number of points. The voltage across the $5000 \mu \mathrm{~F}$ capacitor (two $2,500 \mu \mathrm{~F}$ capacitors, connected in parallel, were used in the prototype) on no load should be approximately 29 volts with a voltage on load of 25 V at 1.5 A . The emitter voltage of TR1 is 13 volts, with 18.5 volts at the collector for an output voltage of 17 volts.
The voltage on the slider of the preset potentiometer is 14 volts. This component should be of high quality as severe damage could be caused by a faulty condition. After experiencing some difficulties with cheap skeleton potentiometers the author decided to use a multi-turn linear track potentiometer which is probably at the other extreme on cost.

## PRE-AMPLIFIER CIRCUIT

The pre-amplinier circuit is shown in Fig. 2.4 and consists of three stages. The input impedance is very low, and allows an unscreened signal interconnection between the pitch boards.

Perspective view of power supply module

It is essential that the amplifier should be able to handle a wide range of signals, since the peak level for one note is approximately 80 mV peak-topeak on maximum attack. The maximum input signal which can be presented at the input is approx. 2.6 volts peak-to-peak which represents 32 notes in the maximum attack position.



Fig. 2.2. Etching and drilling details for p.s.u. printed circuit board (full size)


Fig. 2.3. Component assembly on p.s.u. printed circuit board

COMPONENTS . . .

## POWER SUPPLY UNIT

## Resistors

| R1 | $1 \mathrm{k} \Omega$ | R4 | $1 \mathrm{k} \Omega$ |
| :--- | :--- | :--- | :--- |
| R2 | $100 \Omega$ | R5 | $560 \Omega$ |
| R3 | $100 \Omega$ | R | $3 \cdot 3 \mathrm{k} \Omega$ |

All $5 \% \frac{1}{3} \mathrm{~W}$ carbon

## Potentiometer

VR1 $1 \mathrm{k} \Omega$ preset

## Capacitors

| C1 | $0.02 \mu \mathrm{~F}$ polyester 125 V | C 4 | $100 \mu \mathrm{~F}$ elect. 15 V |
| :--- | :--- | :--- | :--- |
| C2 | $5,000 \mu \mathrm{~F}$ elect. 35 V | C 5 | $50 \mu \mathrm{~F}$ elect. 25 V |
| C3 | $25 \mu \mathrm{~F}$ elect. 15 V |  |  |

## Diodes

| D1-D4 | ZS271 (Ferranti) 4 off |
| :--- | :--- |
| D5 | ZS170 (Ferranti) |
| D6-D7 | K5068A (Ferranti) $6.8 \mathrm{~V} \quad 400 \mathrm{~mW}$ Zener |
|  | $(2$ off) $)$ |

Transistors
TR1 ZTX 300 Ferranti TR3 ZT 3055 Ferranti TR2 ZT 3053 Ferranti

## Transformer

T1 Mains transformer, Douglas type MT3AT Miscellaneous

FS1 - 200 mA fuse, FS2 - 1.5 A fuse, S 1 Main toggle switch, heatsink; LP1 - Mains neon

The output from the first stage is fed via a resistor network to a compound emitter follower containing two transistors (TR5, TR6). The design of this portion of the circuit is determined by the condition required for the tremolo effect. The tremolo circuit, which is described later, acts as a shunt modulator by presenting a variable impedance across VR2, so giving a variable degree of attenuation of the signal between the first and second stages.

Output from the emitter follower is again at low impedance and the nominal attenuation (tremolo off)
for the second stage is a factor of two. Thus the overall voltage gain to the output of the emitter follower is approximately 1.5 .

## LEVEL CONTROL SETTINGS

The second stage is followed by the two level determining elements, the first being the foot soft pedal control, whilst the second is the keyboard level control. The latter item consists of a five position switch with fixed resistors giving stepped attenuation levels.


Fig. 2.4. Circuit diagram of pre-amplifier


Fig. 2.5. Circuit diagram of tremolo unit, S3 is the keyboard on/off control switch


Fig. 2.6. Etching and drilling details for pre-amplifier and tremolo printed circuit board (full size). The white circles for JK1, VR2, VR4 and VR5 should be only $\frac{3}{8}$ in diameter

COMPONENTS . . .

## PRE-AMPLIFIER

Resistors

| R7 | 1k $\Omega$ | R18 | $270 \Omega$ |
| :---: | :---: | :---: | :---: |
| R8 | 2.2k S | R19 | $12 \mathrm{k} \Omega$ |
| R9 | 10k S | R20 | $22 \mathrm{k} \Omega$ |
| R10 | 1.8 k S 2 | R21 | $18 \mathrm{k} \Omega$ |
| R11 | $270 \Omega$ | R22 | $270 \Omega$ |
| R12 | $27 \mathrm{k} \Omega$ | R23 | $6.8 \mathrm{k} \Omega$ |
| R13 | 100k $\Omega$ | R24 | $1 \mathrm{k} \Omega$ |
| R14 | 150k $\Omega$ | R25 | $470 \Omega$ |
| R15 | $1 \mathrm{k} \Omega$ | R26 | $220 \Omega$ |
| R16 | 820k $\Omega$ | R27 | $220 \Omega$ |
| R17 | $3.9 \mathrm{k} \Omega$ |  |  |

## Capacitors

C6 470 pF polyester 125 V
C7 $400 \mu \mathrm{~F}$ elect. 15 V
C8-C9 $1 \mu \mathrm{~F}$ elect. 15 V (2 off)
$\mathrm{C} 10-\mathrm{C} 1230 \mu \mathrm{~F}$ elect. 15 V (3 off)
Transistors
TR4-TR7 ZTX300 (Ferranti) (4 off)

## Potentiometers

VR2 $50 \mathrm{k} \Omega$ carbon lin.
VR3 $2 k \Omega$ carbon log.
VR4 $10 \mathrm{k} \Omega$ carbon log.

## Miscellaneous

S2-5-way single pole rotary switch
JKI-Standard jack socket

## TREMOLO UNIT

## Resistors

| R28 | $1.8 \mathrm{k} \Omega$ | R34 | $10 \mathrm{k} \Omega$ |
| :--- | :--- | :--- | :--- |
| R29 | $100 \mathrm{k} \Omega$ | R35 | $10 \mathrm{k} \Omega$ |
| R30 | $100 \mathrm{k} \Omega$ | R36 | $5.6 \mathrm{k} \Omega$ |
| R31 | $3.9 \mathrm{k} \Omega$ | R37 | $270 \Omega$ |
| R32 | $1.8 \mathrm{k} \Omega$ | R38 | $270 \Omega$ |
| R33 | $10 \mathrm{k} \Omega$ |  |  |
| All resistors $\pm 5 \% \frac{1}{3} W$ | carbon |  |  |

## Capacitors

C12 $0.047 \mu \mathrm{~F}$ polyester 125 V
C13 $0.47 \mu \mathrm{~F}$ polyester 125 V
C14 $10 \mu \mathrm{~F}$ elec. 15 V
C15 $2 \mu \mathrm{~F}$ elect. 15 V
C16 $2 \mu \mathrm{~F}$ elect. 15 V
C17 $100 \mu \mathrm{~F}$ elect. 25 V
C18 $0.47 \mu \mathrm{~F}$ polyester 160 V

## Transistors

TR8-TR10 ZTX 300 (Ferranti) (3 off)

## Potentiometers

VR5 100 k carbon lin.
VR6 $10 \mathrm{k} \Omega$ preset

## Switch

S3-On/off switch

Normal setting is position 2 which gives the nominal output, on maximum master volume, for a five note input. Position 1 gives the nominal output when a single note is played and is only for special single note solo effects.

Positions 3, 4 and 5 give quick operation attenuated levels ais required. In position 2 the attenuation between stages two and three is a factor of five.

The single transistor third stage has a gain of approximately ten, and again has a high dynamic range, such that the maximum output swing capability is 8 V peak-to-peak.

Overall pre-amplifier voltage gain with the level control in position 2 totals a factor of three, and is followed by a master volume control which feeds the external amplifier jack socket, and a fixed attenuation network prior to the main internal amplifier.

## TREMOLO CIRCUIT

The tremolo circuit is shown in Fig. 2.5. and consists of a multivibrator (TR8, TR9) supplying a square wave output across a preset potentiometer in the range of $5-10 \mathrm{~Hz}$. The output from the preset feeds two low pass filters to give a ramp waveform at the base of TR10.

Preset adjustment ensures that the transistor operates from the off condition and provides maximum voltage change at the collector. In turn the degree of voltage swing is controlled by VR2 when applied to the pre-amplifier.

## PRE-AMPLIFIER AND TREMOLO ASSEMBLY

The pre-amplifier and tremolo circuits are mounted on the same printed circuit board, which


Photograph of the completed pre-amplifier and tremolo board
also acts as the panel to hold the external amplifier output socket, the master volume control, and the tremolo rate and depth controls.
Figs. 2.6 and 2.7 give details of the printed circuit board and component assembly. As before all holes should first be checked against the diagrams, and any missed holes should be drilled before proceeding. This may be particularly necessary in the case of the potentiometer peg location holes which will vary in position for different types and sizes of potentiometer.


Fig. 2.7. Component assémbly detail for pre-amplifier and tremolo unit

Next the terminal pins should be fitted, followed by the potentiometers, jack socket, and preset potentiometer. Fit the resistors, transistors and capacitors in that order, and finally wire the potentiometers and output socket.

## MAIN AMPLIFIER

The circuit of the main amplifier is shown in Fig. 2.8 and is based on the Sinclair integrated circuit type IC-12.

The amplifier is intended as a low power unit to drive the two small internal loudspeakers, or the headphones. Whilst the power is adequate for home use in a small or medium size room, an external amplifier is recommended for any serious use of the instrument outside the domestic circle.

The circuit within the package is fully explained in the Sinclair booklet which is supplied with the unit. If you should purchase a unit which includes the earlier simple two page leaflet the booklet can be obtained by writing to Sinclair Radionics Limited.

## COMPONENT CHANGES

A printed circuit card is supplied with each unit and the photograph shows the assembled unit. Some of the component values have been changed from those suggested by Sinclair for the average application. The components affected are marked with an asterisk in Fig. 2.8 and the associated components list.
$\mathrm{C} 19(0.5 \mu \mathrm{~F})$ has been chosen for the input coupling capacitor since, as explained later, the input impedance has been reduced such that bass response could suffer if this capacitor were not increased. $\mathrm{C} 20(0.01 \mu \mathrm{~F})$ has been used to provide a small amount of treble cut as a form of tone shaping on the internal amplifier setting. C25 $(400 \mu \mathrm{~F})$ prevents any possible instabilities due to the length of the supply. lines.

## MAXIMUM BASS

To provide maximum bass response a $1,000 \mu \mathrm{~F}$ coupling capacitor (C27) is connected at the output. Due to the large size of this capacitor it has been removed from the p.c.b. as supplied by Sinclair. It will also be noted that the connection of this capacitor within the circuit has been removed from the ground side of the speakers and inserted on the live side. This prevents the possible shorting out of C27 to earth by the insertion of a jack plug into the headphone socket where the jack plug is connected to some other equipment which is earthed.

This rearrangement means, that is is necessary to connect a wire link across the connections intended for C27, thus making pin F an earth point.

The input impedance is reduced by the choice of R39 which minimises pick-up between the preamplifier board and main amplifier. R41 (470 ) is used to hold the amplifier gain at 60 . This gives an input sensitivity of approximately 200 mV peak-topeak for a 3 watt output into 6 ohms.

Sinclair Radionics state that it is very important that the i.c. only be used mounted on the printed circuit board which is supplied with it. Due to the very high gain of the device, the slightest coupling between output and input can cause trouble in the way of distortion and instability.

## Next month : Pitch Board



Proprietary board assembly for the main amplifier


Fig. 2.8. Circuit diagram of main amplifier

## COMPONENTS . . .

## MAIN AMPLIFIER

Resistors'
R39* $27 \mathrm{k} \Omega$
R40 $150 \mathrm{k} \Omega$
R41* $470 \Omega$
R42 $27 \mathrm{k} \Omega$
R43 $22 \Omega$
All resistors $\pm 5 \% \frac{1}{3} \mathrm{~W}$ carbon

## Capacitors

C19* $0.5 \mu \mathrm{~F}$ polyester 125 V
C20* $0.01 \mu \mathrm{~F}$ polyester 125 V
C21 $10 \mu \mathrm{~F}$ elect. 15 V
C22 $100 \mu \mathrm{~F}$ elect. 15 V
C23 $2,200 \mathrm{pF}$ polyester 125 V
C24 470 pF polyester 125 V
C25* $400 \mu \mathrm{~F}$ elect. 25 V
C26 $0.01 \mu \mathrm{~F}$ polyester 125 V
C27 $1,000 \mu \mathrm{~F}$ elect. 15 V
Integrated Circuit
ICl-Super $1 \mathrm{C}-12$ with printed circuit board. (Sinclair Radionics Ltd.).

## Miscellaneous

JK2-Jack socket to match headphones
LS1-LS2 $3 \Omega$ elliptical loudspeakers $8 \mathrm{in} \times 2 \frac{1}{2} \mathrm{in}$.
S4-On/off switch


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## KEYBOARD LOGIC AND HARDWARE

Satisfactory functioning of the keyboard is vital for correct calculator operation and to make things as easy as possible for the user, a great deal of thoughtful design has been carried out as regards key-top labels, keyboard layouts, the required function of each of the keys, and compensation for possible "finger-trouble"! The author's solutions to these problems are embodied in the Digi-Cal keyboard, but should any constructor wish to alter either the physical layout or the logic there is plenty of scope for modifications to produce a "custom", design.,

## KEYBOARD DEFINITION

The keyboard circuitry can be split up into a number of independent or interlocking sub-circuits, and each of these performs a different function, corresponding to one of the following list of keyboard requirements:

1. Convert figure key depressions into a four bit B.C.D. code and generate a single coincident clock pulse to enter the B.C.D. data into the (separate) entry register.
2. Convert arithmetic key depressions into ${ }^{-}$a static two bit code suitable for use by the programme and arithmetic sections.
3. Record the number of decimal places in each figure entry and control the final alignment of the number so that it has the correct number of decimal places as entered on the thumb-wheel switch.
4. Produce a three bit code representing the number of decimal places to be displayed, dependent upon whether entries or answers are being read out.
5. Generate a single pulse to start the programme when the equals key is pressed.
6. Switch on the overflow lamp when an overFiow. input is received from the arithmetic section.

## BLOCK DIAGRAM

The circuit of the keyboard panel can be subdivided into a number of relatively straightforward logic groups.

Fig. 4.1 shows these groups laid out in the form of a functional block diagram and it is best to understand the keyboard using this diagram before studying the full circuit in detail.
In the block diagram the circuit has been simplified as far as possible and the symbol consisting of a switch in a box with an arrowed input is used to represent gating functions which control the switched path, the arrow showing the input which has control.

## ARITHMETIC KEYS

Starting at the top of the diagram, the four main arithmetic keys are fed to an encoding circuit which generates a unique binary code for each depression. This code is "staticised" in two latch flip-flops so that the code remains on the output wires until overridden by a future (different) arithmetic selection.

The $\mathrm{E}^{2}$ key is in parallel with the "mUltiply" key since squaring is simply a modified multiplication sequence; a separate $\mathrm{E}^{2}$ output is also produced to modify the programme.

The narrow START pulse required by the programme is generated by a monostable triggered by the equals key, and this monostable is also used to start the second half of the multiply and divide sequences on command from the arithmetic section.

## NUMERAL KEYS

The number keys are encoded to their B.C.D. equivalent in a diode matrix, the outputs of which are also staticised by four flip-flops so as to present a steady logic output during the clock pulse which enters the data into the entry register, the data in the latches remaining until either a different number key is pressed, or the equals key is pressed in which

## KEYBOARD LOGIC



Fig. 4.1. Simplified block diagram of the keyboard logic. The symbol of a switch in a box with an arrowed input represents a group of logic gates, the arrow being the input which has control


Fig. 4.2. The complete circuit diagram of the keyboard logic panel. All i.c.s have pin 7 as ground and pin 14 as 5 V except IC14 whose power supplies are shown


Fig. 4.3. A more detailed diagram of the clock pulse gating circuit
case the latches are cleared. An output is also generated from the matrix whenever any key is pressed, and this signal is used to trigger the clock pulse generator, but more of this later.

## DECIMAL POINT POSITIONING

If the decimal point key is pressed during an entry then a latch is set which in turn opens a gate to allow any subsequent clock pulses through to a counter. This counter counts the clock pulses after the point entry and gives an output which is compared with the setting of the decimal point thumbwheel in a comparator circuit.

When the correct number of decimal places have been entered the comparator registers the equality and shuts off the clock pulses from the clock pulse generator. It is this mechanism which prevents entry of too many decimal places in any figure entry.

If the number entered has too few decimal places then when the programme is started a normalise command from the programme allows the main calculator clock pulses through to the clock generator in the keyboard, and the entry register data is rapidly shifted along until the comparator again registers equality then shuts off the clock.

This process ensures that all numbers can be entered without regard to the position of the point, normalisation being automatically carried out by the calculator itself.

## OTHER FUNCTIONS

There are two additions to the clocking circuit to allow for special conditions. The equals key is allowed to close the path connecting the clock pulses to the counter, this being done to cater for the occasions when only whole numbers are entered, the decimal point being unused.

Also, the recall constant ( K ) key is allowed to shut off the clock path to the entry register so as to inhibit normalisation, this being necessary because the constant in store is correctly aligned to start with.

The three bit output of the thumbwheel switch is used to deline decimal point position when the answer to at calculation is being displayed, and the three bit output of the counter is used to define it when an entry is being displayed. Switching between the two sources is carried out by a gating arrangement controlled primarily by the programme.

Any keys not shown on the simplified diagram have no associated logic circuitry and are passed out to the rest of Digi-Cal as switch closures to earth.

The full circuit of the keyboard is given in Fig. 4.2, and with the knowledge of its operation gained from the simplified diagram it should be possible to pick out the various logic groups previously discussed. One particular group may still be a little difficult to trace out, and that is the clock generattion and gating.

To see this circuit in a clearer light it has been redrawn in Fig .4 .3 in an integrated way, and because this section is so important we can go through it step by step.

## CLOCK CIRCUIT

The clocking is initiated by detecting any output from the diode matrix which is an indication of a key depression. Because the normally high A, B. C. and $D$ outputs of the matrix will always have either their true or their complement ontput active (i.e. low) for any one of the ten relevant key depressions a detection circuit can be made by "ORing" together the true and complement version of any one of them.

In Fig. 4.3 Gl performs the detection by using the two " $A$ " outputs from the matrix. The "logic 1 "


The upper side of the keyboard showing the arrangement of the switches in the slots
output from G1 is applied to G2 which is an open collector gate with $R_{1}$ and $C_{1}$ in its collector circuit to give a slow " 0 " to " 1 " transition. The other input to G2 is an inhibiting input from the equals key which prevents a clock pulse being generated by the latch reset when the equals key is pressed.
The output of Gl is shown for a typical key depression and as you can see it carries with it a faithful (though inverted), representation of the switch bounce which is a feature of all mechanical s'witch operations, and which must be removed before the signal is used to trigger the monostable.
The output of G2 is "cleaned-up" by $R_{1}$ and $C_{1}$ to give the output pulse shown, and the slow transition after the key is released is used to activate the level sensitive trigger input of the SN74121. The fairly long negative going pulse from the monostable is "ORed" with the possible output of G3 and fed out to the entry register via G5 which is controlled by an inhibitory input from the comparator equal latch. Gate G3 is used to control the main clock input used during the normalising cycle, being enabled by a long pulse from the programme.

## DIODE MATRIX

The diode matrix used to encode the figure keys into their binary equivalent is so named because of its physical construction, the logic it performs being simply that of eight multi-input or gates, the number of inputs to each gate being set by the number of diodes connected to its output line.

In the circuit diagram the eight horizontal lines represent the gate outputs, the requisite inputs being made by the diode links to the eleven vertical lines from the number keys and the reset line.

The matrix is built using the same technique as was employed for the display matrix (Part 3) with a small piece of Veroboard providing the vertical output lines to the latches and the key inputs being wired to the horizontal strips on the mother board.

Remember that it is necessary to make guide funnels on the blank side of the small piece of Veroboard to ease the location of the diodes prewired to the mother board, and most important, the cathode end of the diodes should be mounted next to the mother board.

The outputs from the matrix drive the preset and clear inputs of the four latch bistables, which do not use their d inputs or clock inputs in this application (IC27, IC28).

The use of SN7474 flip-flops in this position is a bit extravagant, since a couple of SN7400s cross coupled to form four latches would work just as
well and can be incorporated if desired, the more versatile SN7474s being used in the prototype toallow for possible future modification to enter data automatically from a source such as a tape reader.

## FUNCTION CODER

Unlike the number coder and latches, the function coder and latches do not need a separate set of coding gates, the latches being built from SN7410 three input NAND gates, ICs 23 and 24, and the two inputs to each gate left after making the latch coupling are sufficient to provide the gating required to encode the arithmetic keys.
The $\mathrm{E}^{2}$ key is coded in the same way as multiply and a diode D103 is used to facilitate this.

## EQUALS KEY

Pressing the equals key is the signal to start the selected arithmetic programme, and as may be expected, when it is pressed several things happen at once.

The earth appearing on the equals line:
(a) resets the number latches via the matrix;
(b) inhibits the generation of a clock pulse which would normally result from the previous reset (see G2, Fig. 4.3);
(c) sets the decimal point latch IC22 to allow subsequent clock pulses through to the counter (this latch may have already been set by operation of the decimal point key during figure entry).
When the key is released and the earth disappears the resulting positive going signal on the line is "debounced" and used to trigger the start programme monostable IC26 via a simple diode gate D101, D102. The output from the monostable is a very narrow pulse which is used to initiate the programme sequence. The diode gate is required so that the programme can also trigger the monostable to generate a "restart" pulse required in some arithmetic sequences.

## DECIMAL POINT NORMALISATION

When the decimal point latch is set clock pulses are allowed through to the SN7490 decade counter IC14, which in effect counts the number of figures after the point in a number entry. The contents of this counter are continually compared with the decimal point switch output so as to detect a state of equality.

The thumbwheel switch output is already in a binary code though of an inverted form so that a logic " 1 " is represented by an earth connection and

## COMPONENTS . . .

| KEYBOARD PANEL |  |  |
| :---: | :---: | :---: |
| Resistors |  | Capacitors |
| R37, R38 | $5 \cdot 6 \mathrm{k} \Omega \frac{1}{4} \mathrm{~W} \pm 10 \%$ carbon | C7, C9 $22 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. (2 off) |
|  |  | C8, C13 $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. (2 off) |
| Integrated | Circuits | $\mathrm{C} 10, \mathrm{C} 11, \mathrm{C} 12 \quad 0.047 \mu \mathrm{~F}$ (3 off) |


| Integrated | Circuits |
| :--- | :--- |
| IC13 | SN7404 |
| IC14 | SN7490 |
| IC15 | SN7486 |
| IC16 | SN7420 |
| IC17 | SN7410 |
| IC18, IC19 | SN7450 (or SN7451) (2 off) |
| IC20 | SN7401 |
| IC21 | SN74121 |
| IC22, 23, 24, 25 SN7410 (4 off) |  |
| IC26 | SN74121 |
| IC27, IC28 | SN7474 (2 off) |

## Switches

S2 Birch Stolec EB10N 1248 and a pair of end plates
S3-S23 On/off push button (Bulgin type MP22, white 12 off, black 9 off)

Diodes
D57-D103 West Hyde type "red" (47 off)

## Lamp and holder

LP9 6 V 40 mA lamp and holder (Bulgin type D22)

## Miscellaneous

9.5 in $\times 3.3$ in 0.1 in matrix Veroboard



Fig. 4.4. Layout of the components on the Veroboard panel and the cuts which must be made in the copper strips. The links shown are for the power supply lines only and can either be made above or below the board. The black dots on the "diode matrix" board show the positions of the diodes which are arranged with the cathodes towards the main panel


A photograph of the keyboard logic panel in the prototype. The edge connector is an optional extra as wires can be soldered directly to the board
logic " 0 " by an open circuit. This suits our purpose admirably since the SN7486 used as a comparator, IC15, gives a " 1 " from each of its gates when its inputs are exactly opposite.

By comparing the true counter outputs with the complement switch outputs a comparator is formed which gives three "I" outputs when the count equals the switch setting. These three ones are used to enable a gate IC16, which in turn sets a latch, IC17, the output of which forms the input to G5 of Fig. $4 . j$ and serves to shut off the supply of clock pulses to thelentry register.

## DISPLAY CONNECTIONS

The position of the decimal point in the display can be governed either by the contents of the counter, IC14, or by the thumbwheel switch setting depending on a command from the programme.

Switching between these two sources is carried out by the and Or invert gates IC18 and IC19, but note that for this purpose the outputs from the thumbwheel switch are inverted by IC 13 , to make them compatible with the counter outputs. The selected point data is fed to the display via a three line bus.

## OVERFLOW INDICATION

If an overflow condition is detected during an arithmetic operation a " 1 " input on the overflow line drives three SN7401, IC20, gates in parallel to switch on the panel indicator bulb. The correct procedure is then to use the clear key to erase all data and then re-enter the problem in a rescaled form. It is necessary to parallel connect three gates for this job because of the current requirements of the bulb.

## PROGRAMME RESET

To return the keyboard to its correct state for a further calculation it is necessary to rest the decimal POINT and equals latches and the decimal point position counter. This is carried out by a Clear ENTRY command from the programme which, as its name implies, also clears the contents of the entry register.

## CONSTRUCTION

The layout of the Veroboard panel is shown in Fig. 4.4, the panel itself being cut to size from a West Hyde type 122 board.


Fig. 4.5. The switches are wired as shown. All wires except those with an asterisk go to the panel shown in Fig. 4.4

The finished panel will be mounted upside down under the keyboard proper, affixed to the recessed wooden blocks either by means of wood screws, or, as in the prototype, by captive 6B.A. bolts cemented into the blocks with Arâldite.
Needless to say, it is advisable to make all the board fixing arrangements before commencing the wiring.
Wiring up is carried out as before with thin single core wire, using terminal pins where necessary to increase the wire handling capacity of a particular pad.

Off board wiring is achieved with thin flexible stranded wire left long enough that the board can be easily removed for access, but not too long that stray pick-up could become a problem.

An edge connector was used to carry some of the off-board wiring in the prototype to help with development, this would not be necessary, or even advantageous to any significant extent with later versions, due to the very large number of interconnections required in all.

The switches are wired up as shown in Fig. 4.5

## TESTING

As with the display board last month it is advisable to wire up the keyboard logic a section at a time, testing each block before proceeding. In this way small mistakes can be easily discovered and dealt with before they become buried in several different circuit blocks.

Almost all the logic can be tested at the wiring stage without undue complication, outputs being monitored with a multimeter or by connecting them temporarily to the display board input if desired. If the latter course is chosen remember that the display board requires an inverted B.C.D. input, which while easily accommodating the decimal point outputs and function code outputs, will require temporary connection to the $\overline{\mathrm{Q}}$ rather than the Q outputs of the figure latches for correct display
The clock pulse output is long enough to be detected on a meter which is an advantage for tests of this type, the equals triggered STart pulse is however far too short to be detected in this way but fortunately this part of the circuit is quite simple and can be left unichecked at this stage.

Next month: Entry Register (constant store, entry register and multiplexer)

## Pollits ainsin

## SQUARE WAVE GENERATOR

(Sept. 1972)
In the lower half of Fig. 6, the connections for the switch should read as follows reading from top to bottom:
(a) Top group is SIb numbered $1,2,3, \mathrm{~W}, 4$;
(b) Bottom group is Sla numbered $4,3,2,1$;
(c) SIb wiper should read Sla wiper.

# NEWS BRIEFS 

## TECHNICAL INFORMATION SERVICE

ANEW technical information service, aimed at giving firms. corporate bodies, instructors, students, and others engaged in educational work, advice on correct electrical installation practice according to IE.E. "Regulations" has been established by the National Inspection Council for Electrical Installation Contracting.

The service is confined to technical matters and excludes advice on commercial or trading aspects, industrial or labour relations, and information relating to manufacturers' products.
The aim is to promote a high standard of installation and maintenance safety of electrical wiring and appliances. The subscription fee is $£ 3.50$ per year to cover the services offered. Further information is available from the Council at Trafalgar Buildings, 1 Charing Cross Road, London SW1A 2DT.

## NEW POCKET CALCULATOR

What is claimed to be the world's smallest electronic calculator has just been announced by the British sompany Sinclair Radionícs Ltd. At only just over $\frac{1}{4}$ in thick and $5 \frac{1}{2}$ in $\times 2$ in this can really claim to be a pocket sized calculator.

Known as the Sinclair Executive this calculator embodies many new techniques. Firstly the single MOS chip which forms the calculator consumes a mere 20 mW of power making it possible to run the whole machine from three mercury cells of the type used in hearing aids. This is made possible by switching off the i.c. completely between clock pulses relying on the capacitance of the MOS elements to store the information. Secondly the clock rate is reduced when the calculator is not performing a calculation so as to conserve power.

The battery voltage is just sufficient to drive the display thus eliminating the usual dropper resistors which consume power.

The keyboard was specially developed for the calculator being only 4 mm in depth. The contacts are gold plated and the keys are arranged so that it is virtually impossible to touch two at the same time.

The L.E.D. display has a capacity of eight digits. The calculator can perform all the usual arithmetic functions as well as automatic squaring, reciprocals and can operate in either fixed or floating decimal point mode. It also has a memory for calculations with a predetermined factor.

Although at f 79 this calculator is not the cheapest on the market, it must pose quite a threat to the present Japanese domination of the calculator field.

# report from <br> AUSTRALIA 

## BY J.M.WALDIE

In this report, the main theme will be the audio scene, which again should be of interest to intending migrants and visitors.

Audiophiles here have a very large range of components from which to choose; it is likely that the range encountered in the average store will be greater than in a similar situation in the U.K. The possession of a respectable audio system, at least in the large cities, appears to be the rule rather than the exception, despite the "outdoor life" philosophy of the average Australian.

## LOCAL PRODUCTS

The number of companies producing audio systems is relatively small. There are about six brands of Australian-made amplifiers (including Leak and Philips); only two very simple belt-driven turntables are produced here; I understand that there are only two locally produced reel-to-reel recorders and only one cassette system. There are, of course, several brands of "mid-Fi" modular grams and radiograms, including such brands as Kreisler (division of Philips); AWA (Amalgamated Wireless Australia), Thorn, Pye and HMV.

## IMPORT TRADE

However, there is sufficient industry to warrant relatively high import surcharges-designed to protect local manufacturers (which, in some areas, are non-existant). Despite this, Japanese equipment dominates the retail trade, followed by British, German, Scandinavian and - to an increasing degreeAmerican audio components.

For example, of the 30,000 reel-toreel mains-operated recorders and decks imported in 1970-71, 26,000 came from Japan. Some \$A 2 million worth (F.O.B. values) of amplifiers also arrived in the same period, of which well over half also came from Japan, and approximately one-quarter (in terms of value) from the U.K.

The British, however, have the lion's share of the turntable market: of the 200,000 units imported in the year ending June 1971, 176,000 came from the U.K. This is largely due to the practice of installing Garrard and BSR decks in mass-produced "unitaudio" sy'stems.

There is a very wide range of loud-speakers-both imported and home-grown-available, and the Australian audiophile has a bewildering selection from which to choose. There are three locally-made brands of any consequence, and some very presentable designs have been issued. Japanese cabinets again dominate the imported systems. It is in this area that the Americans have been active, promoting the more expensive brands such as JBL, A.R., Marantz and Advent.

## QUADRAPHONY

Largely due to the Japanese dominance in Australia in the audio field (Japan is one of Australia's largest customers for ore and coal; consequently the trade relations between the companies are maintained at a very close level), the current interest among audiophiles is the idea of 4 -channel sound.

The battle between proponents for the CD-4 discrete system of quadraphonic disc reproduction (JVC/ National/RC.A) and the various matrix methods for obtaining more or less the same result (dominated by Sony's SQ system), is now in full swing.

Records produced by both methods are now available; the record manufacturers appear to be taking sides but showing apparent preference for the .matrix method. it has been announced, for example, that EMI, wIth CBS, have opted for the Sony SQ system, but the decision to market SQ records will depend on demand. This demand will presumably originate from new owners of SQ equipment, but who would buy such equipment without a guaranteed source of quadraphonic records?

This chicken-and-egg situation is likely to delay still further the final decision. On the other side of the fence, aurally speaking, RCA have announced, in America, that eventually all records issued by that particular stable will be "in discrete" quadraphony, a move that must be followed not only here but worldwide. And to confuse matters even more, according to local sources of information, A\&M records are being matrixed to the Sansui QS method (not necessarily compatible with Sony's SQ) and two big German companies, Polydor and Deutsche Grammophon, are opting for a discrete system not necessarily similar to JVC/Nivico.

Record manufacturers I spoke to locally all indicated that the outlook was in such confusion that nobody really knows what is happening next.

## ORIENTAL CUSTOM

Australian audiophiles may have an extensive selection of audio equipment available but they don't grow on trees. For English equipment, the list retail price is likely to be double that in the U.K. For example, a Garrard Zero 100 deck has a list price here of \$A 269 (about £120) which compares with the U.K. list price of £54. On the other hand, the difference in price of Japanese equipment is very much less; a Sansui SD 5000 deck is listed here at \$A 756 ( $£ 345$ ) which compares to $£ 328$ in the U.K.

This, of course, explains to some extent why Japanese equipment is so popular. However, we not only import oriental equipment but, to some extent, their methods of selling; it is usual to "haggle". over prices and the published list prices are something of a myth. Moreover, it is always possible to "trade-in" an article, which might not even be electrical.

## COLOUR TELEVISION

It was recently announced that Australia will "go colour" in March 1975. The system to be adopted will be the 625 -line PAL format and all four Sydney stations (1 national, 3 commercial) are in the process of preparing their studios and transmitting equipment.

The manufacturers, however, are treating the news with some trepidation; on the one hand, it could be regarded as a shot in the arm for the electronics industry, but a real concern is being expressed regarding imported receivers which, at the current import rates, could possibly be landed and sold at prices below those expected for locally-produced sets.

The matter will be discussed with the Tariff Board, who decide the magnitude of import duties applicable to any imports. And there are a number of critics who feel that, at an anticipated price of $£ 500$ for a 25 in set, the sales to Australians, who for the most part regard television as a distraction rather than as a necessary part of life, may just not warrant the large initial outlay.

## NEXT


a survey of the commercial scene $\mathcal{1}$ ORGANS
$\mathcal{I}$ effects units

Jguitars<br>\# SYNTHESISERS

## A.F. SIGNAL GENERATOR

For checking and measuring the characteristics of audio equipment, including frequency response, distortion, power output, signal to noise ratio, and dynamic range, one of the most important aids is an a.f. signal generator.

This instrument provides a sine wave output in the frequency range 19 Hz to 200 kHz in four bands. Square wave pulses from 19 Hz to 20 kHz are also available. Maximum output is just over 1 V r.m.s. and can be attenuated internally.



## SIMPLE C-R BRIDGE

Ever had the frustration of finding a capacitor or resistor whose only fault is that the marked value is indecipherable? Now your problems can be over with this simple C-R Bridge.

Values of resistor or capacitors can be determined simply by turning a pointer until the signal from an audio generator is reduced to a minimum.

There is no meter so the unit is inexpensive and robust. The components are readily available and the instrument is easy to construct.

# PRACTICAL <br> HarTRONIES 



## VALUE ON THE BRIDGE

There was a bumper gathering of 300 delegates at the I.E.E. conference on Advances in Marine Navigational Aids. They came from all over the world to hear and discuss 55 papers.
It was a heady occasion embracing just about every navigation technique all the way from visual aids (the earliest known lighthouse was at Alexandria about 265 B.C.) to such twentieth century exotica as "System of location of balloons, ships or buoys by means of a loworbiting satellite".
For the student of marine electronics the occasion was fascination itself. For the practical mariner, expected to foot the bill for equipment as well as understand and evaluate all the rival and complementary systems available, it was more like a nightmare.

The Chief Marine Superintendent of the B.P. Tanker Company Ltd., Mr R. Maybourn, put his finger on the shipowner's problem in his paper "Cost benefit and related factors-a shipowner's point of view'. A navigation outfit for a modern tanker has an installed cost of almost $£ 45,000$ using fairly conventional equipment. Using some of the sophisticated systems described at the conference could result in a doubling of the cost with only dubious advantages.

## EXOTIC IDEAS

Mr Maybourn criticised the electronics industry for developing expensive and exotic ideas without adequate reference to the shipowners, and which are not clearly related to adequately defined problems or needs. He suggested that we often seem to have the solution first and the problem becomes one of finding a need to which to match it.

On computerised anti-collision systems, he wondered whether the problem had been adequately identified or even if the current problem is being solved. He referred to the computer as something which mesmerises the electronics engineer "itching to use its spare capacity in much the same way as an idle hand does a work study practitioner".

What Mr Maybourn asked of the industry was for far more reliability in existing equipment, and some element of standard design in equipment in regard to dimensions, methods of fitting and presentation of data which, he said, was chaotic and certainly would not be tolerated in the aircraft industry.

This point of view deserves some sympathy but should be examined in an historical perspective. Most of the exciting new concepts such as satellite navigation and the v.l.f. Omega system have their origin in military developments as, indeed, did marine radar and Decca Navigator before them. These last two items are now accepted as standard fit and comparatively inexpensive aids. What Mr Maybourn and the shipping industry are getting is the best of both worlds.

## high reliability

If the electronics industry demonstrated anything at all at the Conference it was that today's technology is available with high reliability and at a bargain price, while tomorrow's technology is being seriously considered and will be available in due course to merchant shipping at a fair price.

Timing is terribly important. The Marconi "Predictor" anti-collision radar system has been available for four years, yet less than a dozen have been sold. It was, and is, a brilliant example of electronic engineering and necessarily expensive. Decca Radar brought out a simple anti-collision radar with limited capability about the same time at about half the price and have sold some 700 equipments. It was what the shipowner wanted at a price he could afford. But the outlook for "Predictor" remains bright and Marconi still expect it to do well. It was only the timing that was seriously adrift.

## TOUGH COMPETITION

On the other hand, you don't need to invent new lechnology to invent new systems. I was at the inauguration of the Southampton Harbour Radar system, said to be the finest in Europe if not in the whole world. I searched in vain for completely new technology in both the operations room and the equipment areas. Decca Radar, the prime contractors, had done a wonderful systems engineering job
using off-the-shelf equipment freely available on the open market.

I was also at the unveiling of a new marine radar display concept introduced by Kelvin Hughes and reviewed at the Conference by marine consultant Captain J. P. Stewart. The system uses television techniques and is a notable advance, yet the only device in anv way unusual for a radar is a lightsensitive "image retaining panel". This panel has been commercially available for ten years but nobody had apparently thought of using it in a radar application until Kelvin Hughes originated the idea.

What is worrying the marine electronics industry is an unrealistic pricing policy. Competition is so severe that few companies are making sufficient money to support future civil (as distinct from naval) $R$ and $D$ at a reasonable level.

## TXEZ TAKES TO <br> THE ROAD

The Post Office is always fair game for the knockers and I have knocked away in my time with the best of 'em. SDS Components Ltd, the Portsmouth-based component distributors have gone one step further than knocking by abandoning postal delivery and employing Securicor to make all deliveries.

But the Post Office is not all that bad, although you might still think so when out driving and you get stuck behind one of the eighty or so 27 ft long trailers now being built to house mobile electronic telephone exchanges. Each of these will house a Plessey TXE2 equipment and will be on the move to a site where 1,000 would-be telephone subscribers can stop biting their nails with frustration and get hooked into the network.

The Post Office already has 250 mobile exchanges of conventional types but these can onlv handle 400 lines each. The TXE2 electronic exchange in two trailers can handle 1,000 and a dual installation housed in three trailers. will serve 2,000 subscribers.

Deliveries commence next Mav and the investment by the Post Office is $£ 3$ million. At least they are trying to keep pace with a demand that seems insatiable.

## GOONHILLY 3

I was sorry to see that the 10th anniversary of the Post Office satellite station at Goonhillv Downs passed with so little notice. The occasion was marked by the inauguration of Goonhilly 3 making the station the busiest satellite terminal in Europe.

It seems more than ten years since we were all startled by Telstar 1 and Relay before Early Bird, the first of the intelsat satellites, really got going with a full commercial service.

## Build yourselfa TBANSISTOR RADIO

 NEW! ROAMER 10 WITH VHF INCLUDING AIRCRAFT10 transistors. 9 tunable wavebands. mwl, Mwh. LW. SWI, swr, sws, trawler band, vef and local stations and aircraft band.
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Easy to follow instructions and diagrams. Parts price list and easy build plans 30p (FREE
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7 TUNABLE WAVEBANDS: MW1, MW2, LW, 8W1, SW\&, SW3 AND TRAWLER BAND. Built in ferrite rod aerial for MW and LX. Retractable chrome plated tele600 mW transistors. Car aerial and tape record socket. Selectivity switch. Switchel earplece socket complete Hith carpiece. 8 transistors plus 3 diodes. Fine tone moving coil speaker. Air spaced ganged turing condenser. Volume/onjoff, tuning. wave change and tone controls. At tractivecase in rich chest nut shade with gold blocking. Size 9 in $\times$ in $\times 4$ in approx. Easy to follow instructions and diagrama. Parts price list and easy build plans 2.ap (FREE with parts). P. P. \& INS, 41P TOTAL £6.98 (OVERSEAS
P. \& P, \&I)

POCKET FIVE

3 TUNABLE WAVE-
BANDS: MW, LW,
TRAWLER BARD
MW BAND FOR EASIER TUNING OF LUXEMBOURG, ETC. 7 stages- 5 transigtors and 22 diodes, supersensitive ferrite rod aerial, fine tore moving coil speaker. Attractive black and gold case. Size $5 \frac{1}{3}$ in $\times$ 1 in $\times 3$ in. Easy build plans and parts price list
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TOTAL
BUILDING COSTS $\begin{aligned} & \text { P. P. \& INS. } 21 p \\ & \text { (OVERSEAS }\end{aligned}$
BUILDING COSTS

## NEW! "EDU-KIT"

BUILD RADIOS, AMPLIFIERS, ETC., FROM EASY STAGE DIAGRAMS. FIVE UNITS INCLDDINGMASTER ONIT TO CONSTRUCT, Component include: Tuning Condenser: 2 Volume Controls: 2 Ferrite Rod Aerial: 3 Plugs and Eocketa; Battery Ferrite Rod Aerial: 3 Plugs and Bocketa: Battery Boards Basanced Armature Unit:
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parta). Earpiece with plug and switcheel socket for



ROAMER SIX
 (1)
${ }^{6}$ tomable


MW. LW,
SWl. SW8.
BAND PLOS AN EXTRA MW BAND FOR EASIER TUNING OF LUXEMBOURG, ETC. Sensitive ferrite rod aerial and telescopic aerial for short waves. 3 in speaker. 8 stages- 6 tranaistors and 2 diodes including micro-alloy R.F. transistors, etc. Attractive black case with red grille, dial and black knobs with polished metal inserts. Size 9 in $\times 5 \frac{1}{2} \times 2$ in approx. Easy build plass and parts price list 15p (FREF Fith private listening 30 p extra. TOTAL


## TRANSONA FIVE <br> 5 TRANSISTORS AND 2 DIODES

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(OVERSEAS
P. \& P. 63p)


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BAND. Sensitive ferrite rod aerial for MW and LW. Telescopic aerial for short waves. 3in peaker. 8 inproved type transistors plus 3 diodes. Attractive polished metal inserts. Size 9 in $\times$ 战in $\times 2$ in approx. Push-pull output. Battery economiser switch for extended battery life. Ample power to drive a larger apeaker. Parts price list and easy build plans 25 p (FREE with parts), Earpiece FIth plug and awitched TOTAL

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MAKE A REV COUNTER FOR YOUR CAR
The 'TACHO BLOCK'. This encapsulated block will turn any accurate rev. counter for any car with normal coil ignition

## £1 each

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FULLY TESTED AND MARKED SEMICONDUCTORS
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| ACl 26 |
| :---: |
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$A C 176$
$A C Y 17$
AF239
AF186
$\begin{array}{ll}\text { AC186 } & 0 \\ \text { BC148 } & 0 . \\ \text { BC154 } & 0.20\end{array}$

| 6p |  |
| :--- | :--- |
| 0.15 | $O C 170$ |
| 0.15 | $O C 171$ |
| 0.17 | $O C 200$ |
| $\mathbf{0 . 1 5}$ | $O C 201$ |
| 0.20 | $2 N 1302-3$ |
| 0.20 | $2 N 1304-5$ |

BC 154
BC 107
BC
$\mathrm{BC1} 107$
BCl 108
BCl
BC109
BF199
BC169
BF194
BF274
BFY50
BF274
BFY50
BSY25
BSY25
BSY26
BSY゙27
BSY27
BSY28
BSY29
BSY29
BSY95A
OC41.
OC41
OC44
$O C 45$

## oc71 OC72

| OC72 |
| :--- |
| OC81 |
| 0 Cl |

$0 \mathrm{OC81D}$
$0 \mathrm{CB3}$
0 Cl 139
QCl40

## F.E.T. PRICE <br> BREAKTHROUGH!!

This field effect transistor is the 2 N 3823 in a plastic encapsulation. coded as 3823 E . It is also an excellent replacement for the 2 N3819. Data sheet supplied with device. 1-10 30p each, $10-50$ 25p each. $50+20 \mathrm{p}$ each.


## A CROSS HATCH GENERATOR FOR $\mathbf{f} \mathbf{3} .50$ : 1 :

YES, a complete kit of parts including Printed Circuit Board. A four position switch gives X-hateh, Dots, Versical or Horizontal lines. Integrated Circuit design for easy construction and rellability. This is a project in the September
This complete kit of parts costs E3.50, post paid.
A MUST for Colour T.V. Alignment.
Our famous PI Pak is still leading in value for money. Full of Short Lead Semiconductors \& Electronic Components, approx. 170. We guarantee at least 30 really high quality factory marked Transistors PNP \& NPN, and a host of Diodes \& Rectifiers mounted on Printed Circuit Panels. Identification Chart supplied to give some information on the Transistors.

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## PRTENTI RETMEM

## MOVEMEHT DETEGTOR



Fig. 1

British Patent 1267194 from Rotameter Manufacturing Co Ltd describes a simple but probábly very effective circuit for sensing and registering very small physical movements of objects.

The basic circuit is shown in Fig. 1 and relies on a Hartley oscillator with iwo magnetically coupled inductors L1, L2, a bridging capacitor C 1 and d.c. blocking capacitors C2, C3. The variable resistor VR1 controls, along with the mutual inductance of inductors L1, L2, the amplitude of the circuit oscillations. The frequency of the oscillations depends upon the values chosen for the inductors L1, L2 and capacitor C1.

A small shutter X 1 , formed from non-magnetic but electrically conductive material (e.g. copper, brass, etc), is movable into and out of a small gap which is left between the magnetically coupled ends of the inductors L1, L2.

If the object whose movement is to be sensed is mechanically coupled to the shutter, then any movement, however slight, will cause a change in the amplitude of oscillation. This change may in some cases be reasonably linear.
By amplifying TR1 output a direct dial readout can be obtained or, alternatively, the amplified output can be fed to a "pot coil" electromagnet arranged to apply a restoring force to the object which has moved. This restoring force will tend to move everything back to the original or null position, i.e. seek equilibrium in this state.
Component values are suggested which will detect and register a 0.025 mm movement with a transis tor output change of 200 millivolts.

## ELECTRONIC SPEEOOMETER

In conventional electronic speedometers there is sometimes a problem in raising sufficient control voltage to not only display the speed, but also record the maximum speed obtained. Mechanival systems can introduce an inaccuracy of around 15 per cent which can be hopeless when the engine speeds are in the order of 10,000 r.p.m.

Porsche KG now claim (BP 1254486 ) a form of electronic speedometer which should be free from such problems. In Fig. 1 Porsche show how pulses derived from the ignition coil (at a frequency directly proportional 10 the engine speed) are shaped, integrated by the filter circuit and fed via the sefvo amplifier to a reversible servo motor. This motor drives the speedometer pointer through gearing to provide a direct readout indication of speed.

The car battery voltage is fed to the stabiliser to give a stabilised output voltage which is fed to the shaper and is also tapped by the servo motor to provide a feedback voltage to the servo amplifier. The servo motor also drives a maximum speed indicator through a ratchet which can be released for zeroing by a spring loaded button.

If it is thought wise, and safe, to provide an automatic switch-off for the engine, e.g., to avoid stripping the car gears, the speed voltage signal at the filter can be compared by the critical speed circuit with a predetermined limiting value to operate an ignition interrupter switch.

Instead of sensing pulses derived from the ignition coil, an
independent inductor associated with the engine shaft can be used to produce the necessary pulses.

## Patent actility in the optical FIBRE FIELD

At least two different uses for light fibres in motor cars have now been patented by Joseph Lucas (Industries) Ltd.

In the first patent, published a few months ago (BP 1258395 ) Lucas described and claimed as novel the use of a light fibre with one end set in the door frame and the other behind a dashboard warning indicator. The door itself carries a mirror which is arranged so that light is reflected from a lamp into the doorframe end of the fibre only when the door is closed. Light thus emerges from the other end of the fibre to illuminate the dashboard indicator when the door is safe.
In their more recent patent (BP 1264325 ) Lucas claim as novel, the use of fibres extending variously from all the crucial car driving lights (headlights, sidelights, etc.) to various dashboard instruments. Thus any unlit instrument on the dashboard is a failsafe signal of an extinguished driving light.

The above patents are only reported briefly, because they are technically simple. But in view of recent published correspondence on this topic they may be interesting references for anyone working in the field. Incidentally, it should be borne in mind that although the patents have only recently been published, they were applied for several years ago.

Fig. 1


# marihet PLACE 

Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

## FIRE ALARM KIT

The new fire regulations require all properties which provide overnight accommodation for more than six people to be fitted with an acceptable fire alarm / system. To meet this need Ramick Security Services have developed a new "Fire Alarm Kit". complete with instruction and wiring diagram to enable the handyman or an electrician to install a system for the smaller size hotel or boarding house.

The kit costs $£ 30$ complete and comprises a Control Unit; Heatswitch; Preswitch; Testswitch and a 12 V d.c. 4 in gong alarm bell.

The control unit operates from the mains a.c. supply but the voltage to the sensors is only 12 V d.c. A battery is incorporated in the unit together with an automatic changeover circuit so that in the event of a mains failure the installation remains operative and changes back to a.c. supply when reconnected.

Whenever the alarm bell rings it continues until the control unit is reset manually. The control unit is contained in a pressed steel case, suitable for wall mounting, and has a mains indicator light and an on/off switch on the front panel.

The Heatswitch is a bi-metallic switch with a combined rate of rise and fixed temperature automatic fire detector to B.S. 3116. The contacts close when the temperature reaches $135^{\circ} \mathrm{F}\left(57^{\circ} \mathrm{C}\right)$ or if there
is an unusally rapid increase in temperature. The unit can be mounted in any position and provides coverage of approximately 500 square feet per heatswitch.

The Preswitch is a push button unit with a cast aluminium front plate and can be supplied for surface mounting or flush mounting. Breaking the glass front operates the button which activates the alarm.

The Testswitch has a manually biased push button and indicator light complete with housing and is designed for fitting at the end of the twin wire circuit connecting the sensors to the control unit. When the switch is operated the indicator light is energised (and remains energised until the control unit is reset) and the control circuit, external wiring and alarm bell is checked.

Further details can be obtained from Ramick Security Services, The Lodge, Randalls Road, Leatherhead, Surrey.

## PRINTED CIRCUIT BOARDS

Sometime ago we mentioned a versatile printed circuit board called Nippiboard marketed by NIP Electronics.
Although these boards have proved very successful, from the response of users they seem to suffer from two limitations. These limitations are that the boards are not pre-drilled and, more important, do not take integrated circuits.
Now, NIP Electronics have introduced a new range of boards called "NIP-E-Boards." These boards are available pre-drilled and are suitable for transistor circuits, integrated circuit designs or a combination of both.
The integrated circuit range of boards were designed in order to cater for the vast expansion of component manufacture in standardised DIL packages, such as relays, readout devices, as well as integrated circuits and transistor arrays.

All boards are made from s.r.b.p. or glass fibre and are available predrilled or undrilled according to choice.

The designs of the boards are such that they will all accept edge connectors; a range of card handles, stand-off mounting pillars, and edge connectors is also available.

Further details and prices of the NIP-E-System can be obtained from NIP Electronics, P.O. Box 11, St. Albans, Herts.

## POCKET METER

Readers should find the TS-60R Multimeter from West Hyde Developments an excellent purchase.

This pocket multimeter makes no pretentions to be up to professional standard but still covers most ranges needed for checking out the majority of circuits published in this magazine.

The meter movement is a $1000 \Omega / \mathrm{V}$ type and the ranges covered are: a.c. voltages $0-15 \mathrm{~V}$, $0-150 \mathrm{~V}, 0-1000 \mathrm{~V}$; d.c. voltages $0-15 \mathrm{~V}, 0-150 \mathrm{~V}, 0-1000 \mathrm{~V}$; d.c. current $1 \mathrm{~mA}, 150 \mathrm{~mA}$; resistance $100 \mathrm{k} \Omega$. A feature of the meter is the "printed circuit" rotary switch.
The meter costs only $£ 2$ and in the event of a meter breakdown it is probably cheaper to throw it away and buy a replacement, where. as with the more expensive counterpart it would cost more to repair than the purchase price of the TS-60R.

Further details of the TS-60R multimeter can be obtained from West Hyde Dèvelopments Ltd., Ryefield Crescent, Northwood Hills, .Northwood, Middlesex.

Whilst on meters, a comprehensive range of Avometer spare parts is always held in stock by Instrument Services and they are able to offer a rapid Avometer repair and recalibration service.

Complete refurbishing to "as new" is claimed to take approximately seven days. The company also offers a similar service to users of G.E.C. "Selectest" and "Minitest" instruments.

Further information of the meter repair service can be obtained from Instrument Services Company (London) Ltd., 208 Maybank Road, London E18.

(above) NIP-E-Boards from NIP Electronics
(left) Ramick Fire Alarm Kit
(right) TS-60R Multimeter from West Hyde


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## Riondoula SEEETON RHOM OUR POSTAAG

Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

## Strong objection

Sir-1 object strongly to the article Digi-Cal. Whilst 1 can understand people building their own electric organ or colour TV, l am sure nobody with any sense will build this machine at $£ 110$. These calculators can be purchased from about $£ 60$ : Further to this, the cost of a simple counter with Nixie tube readout is approx. $£ 5$ per unit, i.e. with a six digit readout this section is approx. $£ 30$. Careful design should therefore give a complete calculator for $£ 60$. Still not worth making.
As a "Practical" magazine the very least I would expect is for articles to be practical propositions.
G. A. Bobker,

Unsworth, Bury, Lancs.
far too involved to be explained to "mere amateurs".

It does seem to be true that calculators are going the way of medium-wave transistor pocket radios, where the vast quantities which can be sold reduce prices to a level where amateur designs are difficult to justify on economic grounds, and to those readers who want a cheap calculator to use immediately without knowing what makes it tick, I can thoroughly recommend the new ready made machines available from retailers at low cost.

For those of us who also want to know what's aoing on inside. II hope Digi-Cal will help. After all, now that single-chip organs, f.m. tuners, and even computers are with us, it may be our last chance.
—R. W. Coles.

I am sorry to hear that Mr Bobker objects so strongly, to the Digi-Cal series, and though in the light of the latest developments in the field of m.o.s. technology I can sympathise with his comments on price differential, his letter does point to one very important advantage of these current articles.
The logic operations involved in any calculator are necessarily complex, and the new "single-chip" devices are valuable simply because the complex logic has been "wired-up" entirely by the i.c. manufacturer. Mr Bobker's concept of a calculator made from SN7490's and Nixies shows, 1 am afraid, a lamentable grasp of the way machines of this type operate, and since he himself is active in the circuit design field he could learn a tremendous amount about logic in general and calculating circuits in particular by following the Digi-Cal series. I am sure-many readers will be glad of the chance to delve into this interesting and relevant field of electronics before the rapid march of technology reduces us all to mindless "wireruppers" of a circuit which can be built in half an hour but which is

## What is the point?

Sir-Over the years 1 have successfully constructed many of the circuits published in P.E. The amount of enjoyment derived from assembling these projects has been twofold. Firstly, the satisfaction of building something with one's hands and seeing it working and secondly, knowing you have saved quite a bit doing it.

Therefore when you announced the P.E. Digi-Cal, which is the sort of project right up my street, I could hardly wait for the August issue to get cracking. Imagine my dismay at seeing the total cost as being approx. $£ 110$ when one can buy a ready built article to the same specification for $£ 43 \cdot 25$.

Even if your design had cost the same I would have built it for the satisfaction to be obtained, but 1 cannot see anyone wanting to construct Digi-Cal with such a price difference. So really, what is the point of publishing a design like that.

This is my only criticism so far of an otherwise excellent magazine which I have taken since it first came out.

Brian Harrison Settle, Yorks.

## Beyond reason

Sir-I have been reading Practical Electronics since its inception in November 1964 and on average consider about 50 per cent of the constructional projects of interest to me (not that that figure relates to the quantity constructed!). But aren't we getting a little above ourselves with the introduction of a project to build an electronic calculator?

Now I am not one to decry the technical capabilities of your readers (quite the opposite) but what astounds me is the cost. Your author quotes a bulk component cost $£ 110$ yet one can buy an equally sophisticated machine (with all the trimmings like leading zero suppression. floating DP, constant key, etc.) for $£ 56$ and no extras like replacing over-soldered transistors or mis-wired ic's!

My point, to which I laboriously come, is what is the function of the constructional feature? I think it should be a medium through which constructors can promulgate their brainchildren so that readers can build a piece of equipment which is impossible to buy either because it is prohibitively expensive or because no manufacturer's specification matches that of the constructors. The intrinsic pleasure of building is enhanced when one notices that there are still pennies in the pocket.

Come on P.E. Continue to disseminate the superb features that you do but let us keep our exuberant competition with the pros within the bounds of reason.
J. E. Barrett, London, S.W.6:
We recognise and understand the bewilderment caused to some readers because of the relatively high cost of building the P.E. Digi-Cal. This very creditable project provides a dramatic illustration of the far-reaching effects of rapid and stupendous advances in technology. See Editorial Comment, page 817.

## Advance information

Sir-Could you please inform me if it is possible to purchase a full copy of Practical Electronics,

Digi-Cal drawings and layout diagrams in advance, so that our apprentices may build as a project this very interesting and useful unit.

We would also like to know if it is possible to extend the readout to ten digits without more complex circuitry.

The advance purchase of the drawings, etc. will in no way effect our desire for obtaining further copies of Practical Electronics every month. As the Instructor in this Apprentice School I advocate all our electrical apprentices to get this magazine as it is so useful to them. For myself 1 have every copy since November 1964 having read ninety per cent of their contents (in-cluding-"Pleased to meet you") and all these have been bound into books year by year.
G. Bolsover,

Apprentice School, Perkins Engines Co., Peterborough.

This is not an uncommon request, but we must explain that it is just not possible to supply either individuals or organisations with full details of constructional projects prior to publication in the magazine. The division of the text for all major projects is carefully planned and the various parts of such articles presented in the most logical sequence to permit the construction work to proceed steadily, month by month.

Regarding your second point, the existing design cannot be readily modified to accommodate the extra i.c.'s and other components required to increase the readout as suggested.

## P.E. Scorpio ignition system

Sir-l have installed the P.E. Scorpio ignition system in a car to which I had previously fitted an impulse tachometer.
The instrument is a "Smiths" for which the fitting instructions refer to models IT4, IT4C and IT6. I have connected it into the lead from the coil to capacitors C6/C7 (tag 5) as suggested in Readout, March issue.

Some of your readers may be interested to know that this appears to work satisfactorily.
H. R. M. Foster,
Beverley,
Yorks.

## Logic identifier

Sir-With reference to Mr T. P. Schofield's "Logic Semiconductor Identifier" in the July issue.

Mr Schofield recommends the use of 7440's as lamp driving logic. Although these have a normalised fan-out of 30 , I would recommend the use of 7407's. The latter are hex. buffer/drivers, capable of sinking 40 mA per'buffer, and their cost per package is less than the average price of one 2 N1305 of which four are required in the equipment as designed.

Fig. I shows the use of a 7407 device in the circuit.
P. Lockerby, Nelson, Lancs.


Fig. 1. Suggested circuit for a logic identifier. Note two pairs of buffers are paralleled solely to save tying unused inputs with resistors

## Live wire

Sir-Mr W. A. Rawcliffe the exceptional caravanner-July Readout) seems to have forgotten that Practical Electronics readers pride themselves in their ingenuity.

Having reached the same conclusion as W. A. R. that some indication was required that the flasher bulbs had not gone open circuit, 1 very simply wound 300 turns of 24 gauge plastic covered wire (about 30 feet) on to an old P.O. 3000 relay former and wired it in series with the extra "live" lead he talks about.

This gives ample magnetism to operate the relay and through it contacts an additional pilot lamp on the dashboard while at the same time having a resistance of less than $\frac{1}{2} \Omega$, drops less than 1 volt in doing so-thus not affecting the brightness of the flasher bulb by any noticeable amount.

This system could also be easily adapted for brake light if anyone should so decide and slightly less turns on the former would indicate when only one brake light has gone open circuit!!
H. G. Hearn,

## Be specific

Sir-l have been reading Practical Electronics for two years now, and find it difficult to obtain many of the parts specified in your constructional projects. This applies particularly to transistors, as a lot of your work is done around "Ferranti" transistors, which I cannot find at reasonable prices in Australia. Equivalents lists are a great help, as most of the transistors 1 want are either not listed, or have a CV equivalent only, or a type which is itself out of production, or otherwise unavailable here.

I think that the publication of the characteristics of the transistors used, specifying any necessary parameters, e.g. power dissipation, maximum and minimum beta, and so on, and then go on to give one or two examples. Perhaps when "or similar" appears, you might like to put a few "or similars".

This practice is not confined to transistors, so please don't specify a specific component, just its characteristics and tolerances on those characteristics.
A. Wright,

New South Wales,
Australia.

## Wigan radio club

Sir-I would be obliged if you could find space in P.E. to publicise the forthcoming formation of the Wigan and District Amateur Radio Society.

A temporary committee has been set up and 1 would like to hear from prospective members.
V. T. Brooks,

8, Sutherland Road,
Worsley Mesnes,
Wigan, Lančs.

## COURSES

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giass printed circuit board $65 \mathrm{p}+10 \mathrm{p}$ P. \& P. Silver-grey stove enamelled glass printed circuit board $65 \mathrm{p}+10 \mathrm{p} P$. \& P . Silver-grey
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RESISTORS
RESISTORS

|  |  |  | ohms |  | + |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| C | 1/20w | 5\% | 82-200K | E12 | 9 | 8 | 7 |
| C | 1/8w | 5\% | 4.7-470K | E24 | 1 | 0.8 | 0.7 |
| C | 1/4W | 10\% | 4.7-10M | E12 | 1 | 0.8 | 0.7 |
| C | 1/2W | 5\% | $4 \cdot 7-10 \mathrm{M}$ | E 24 | $1 \cdot 2$ | 1 | 0.9 |
| C | 1w | 5\% | 4.7-10M | E12 | $2 \cdot 5$ | ? | 1.9 |
| M0 | 1/2W | 2\% | 10-1M | E24 | 4 | 3 | 2 net |
| Ww | 1W | $\begin{aligned} & 10 \% \\ & \pm 1 / 20 \Omega \end{aligned}$ | 0.223 .9 | E12 | 7 | 7 |  |
| WW | 3W | 5\% | 1-10K | E12 | 7 | 7 | 8 |
| WW | 7 W | $5 \%$ | 1-10K | E12 | 9 | 9 | 8 |
| Codes: $\mathbf{C}=$ carbon film ligh stability low noise MO = metal oxide Electrosil TR5 ultra low noise WW = wire wound Plessey |  |  |  |  |  |  |  |
| Values: <br> E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and the ir decades. E24: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, $01,62,72,91$ and their decades. Prices are in pence each for same ohmic value and power rating, NOT mized values. (Ignore fractions of $1 p$ on total ealue of resistor order.) |  |  |  |  |  |  |  |

## Minitron <br> DIGITAL <br> INDICATOR

TYPE 3015F Seven seg. ment indicator compatibe with standard supplies. Figs. 0-9 fros well iltuminated filament

segrnents to give
character of 9 mm height plus decimal point Power requirement 8 ml from $5 V$ d.c. per segment. A limited number of alphabetical symbols also available. In 16 lead DIL case
$\leq 2$

Suistable BCD decoder driver £1.36 type FLLI21T
DIL Socket; 16 lead 30p. No. 3015 G showing + or - and fig. 1 and decimal point 28.

## PRECISION COMPONENTS

Resistors by Rivlin, $0.1 \%$ to $0.01 \%, \ldots 1$ to 52. Capacitors by MFD, $0 \cdot 1 \mu \mathrm{~F}$ to $22 \mu \mathrm{~F}, 70 \mathrm{p}$ to £3.50. Priees and delinery timen on request.

## SLIDE POTENTIOMETERS



Robust construction, smooth silent action In values from $4 k 7 \Omega$ to $1 M \Omega$, ilinear or $\log$ 26p each. Escutcheon, light grey, 10p.
Knobs, tiat, grip type, in hack/red/green 5 p each.

## SIEMENS CAPACITORS

POLYCARBONATE-5\% TOLERANCE
250 V up to $0.1 \mu \mathrm{~F}: 100 \mathrm{~V} / 0-1 \mu \mathrm{~F}$ and above $0.01,0.012,0.015,0.018,0.022,0.027,0.033,0.047 .0 .056$, $4 p$ each.
$0.068,0.082,0.1,0.12,0.15,4 p$ each.
$0.18,0 \cdot 22,5 \mathrm{D}$ esch.
$0.27,0.33,6 p ; 0.39,7 p ; 0.47,8 p ; 0.56,10 p ; 0.68,11 p$ $1 \mu \mathrm{~F}, \mathbf{1 3 p}$.

## ELECTROLYTIC CAPACITOR

## 17 100, (Values in $\mu \mathrm{E} / \mathrm{V}$ )

$0 \cdot 47 / 100,1 / 100,2 \cdot 2 / 63,4 \cdot 7 / 35,10 / 25,22 / 16,47 / 10,47 / 25$ $100 / 10,220 / 3,7 p$ esch.
$10 / 63,22 / 35,47 / 35,100 / 16,100 / 25,220 / 6,220 / 10,220 / 16$, $10 / 63,22 / 35,47$
$470 / 3,7 \mathrm{p}$ each.
$47 / 50,47 / 63,100 / 35,220 / 16,8 \mathrm{p}$ each; $100 / 50,220 / 3 \%$, 10 p each; $100 / 63,470 / 20,1000 / 10$, 12p each; $220 / 63$ $470 / 35,1000 / 16,17 \mathrm{p}$ each; $1000 / 2 \mathrm{c}, 20 \mathrm{p}$ each; $470 / 63$ $1700 / 16,40 \mathrm{p}$ each.
Tantalum and other capacitors, etc.. see latest 1972 catalogue-issue No. 6.

## 1972 ELECTROVALUE CATALOGUE (No. 6)

Third printing with iatest price and information detalla. Contains 96 pages plus cover. More items, more information, more diagrams than ever. Post Ireain U.X. 10 p.

## SOLDERSTAT SOLDER

As appointed distributors for well-known Elremco16 or $24 \pi$ matt, a.c. masine, net $\in 1-87$

INFINITELY VARIABLE,TEMPERATURE CONTROLLED SOLDER IRON $£ 9 \cdot 20$ DE-SOLDER BRAID
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## NIPPIBOARDS

A useful board for circuit building.
Type 1A. Single besic pattern $2 \cdot 2^{\prime \prime} \times 1 \cdot{ }^{\prime} \%$, 15p.

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Many others in stock ug to 10 patterns.

## PUBLICATIONS

Hanibook of Transistor Equivalents, 40p. Handbook Hanibook of Transistor Equivalents, 40p. Handbook
Tested Trangistor Circuits (H. Ness), 40p. Of Tested Tranaistor Circuits (H. Ness), 40p. Colour corles data wall chart, 15 p . Engineera Reference Handbook \& Tables, 20p. (And 3p for postage on each of above if bought separately.)

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| INTEGRATED CIRCUITS |  |  | These prices are uell. |  |
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| FLH 191 (7402) | 20p | FLJ 151 | (7475) |  |
| FLH291 (7403) | 20D | FLJ131 | (7476) | 45 |
| FLH211 (7404) | 25p | FLH221 | (7480) | 68p |
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| FLH351 (7413) | 35 p | FLa 221 | (7491A) | . 128 |
| FLH121 (7420) | 20p | FLJ171 | (7492) | 85 |
| FLHH181 (7430) | ${ }^{2010}$ | FLJJ181 | (7493) | $80$ |
| PLH141 (7440) | 24 1.22 | FLaj231 | (7494) | $\begin{gathered} 801 \\ 1.18 \end{gathered}$ |
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LIC 709C/14 dual in line, 34p. $\quad 709 \mathrm{C} / 5$, TO5 39p LIC $741 \mathrm{C} / 14$ dual in line. 40 p . $\quad 741 \mathrm{C} / 5$, TOJ 42 p .
CARBON TRACK POTENTIOMETERS CARBON TRACK POTENTIOMETERS, long spindles. SINGLE GARG linear 1000 to $2.2 \mathrm{M} \Omega, 12 \mathrm{p}$; Single gang log, $4.7 \mathrm{E} \Omega$ to $2.2 \mathrm{M} \Omega$, 12p; Dual gang linear $4.7 \mathrm{k} \Omega$ to $22 \mathrm{M} \Omega, 42 \mathrm{p}$; Dual gang log, $4.7 \mathrm{k} \Omega$ to 22 Mn . 429; Log/antilog, $10 \mathrm{k} \Omega, 2 \mathrm{~kg} \Omega, 4 \mathrm{k} \Omega, 1 \mathrm{Mn}$ only 42p; Dual antilog, 10k $\Omega$ only, 42p. Any type with 2A D.P. mains owitch, 12 p extra.
Only decades of 10,22 and 47 available in ranges quoted.
DUAL CONCENTRIC in any combination of P. 20 values, 60p; with switch, 72p. Knobs, pair 24p.

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## $15 \%$ on orders over $\& 15$

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## Sinclair Project 60

## Stereo 60



Built and tested post free £9.98

## pre-amplifier/control unit

The versatility of Project 60 high-fidelity modules is well demonstrated in this excellent unit. It provides the facilities essential to good stereo and will enhance the quality of any system it is used with, whether Project 60 or any any other top line power amplifiers. Compact, yet robustly constructed, the unit is easily panel mounted and will operate satisfactorily from 18 to 35 volts supply. Silicon epitaxial transistors are used throughout to achieve a very high signal to noise ratio with excellent separation between channels. Distortion at maximum output is barely $0.02 \%$ with magnetic p.u. input. Accurate equalisation is provided for all inputs, which are selected by push buttons. For maximum effectiveness, the Sinclair A.F.U. is recommended for use with the Stereo 60 pre-amp/control unit. A comprehensive manual supplied with Project 60 modules makes installing and connecting easy and ensures best possible results from your system.

## Super IC. 12

Integrated circuit
high fidelity amplifier


Having introduced Integrated Circuits to hi-fi constructors with the IC.10, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super IC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up. F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible. It is the equivalent of a 22 tran-
sistor circuit contained within a 16 lead DNL package, and the finned heat sink is sufficient for all requirements. The Super IC. 12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Complete with free manual and printed circuit board.

SPECIFICATIONS
Output power: 6 watts RMS continuous (12 watts peak). 6-8 . Frequency Response: 5 Hz to $100 \mathrm{KHz} \pm 1 \mathrm{~dB}$. Total Harmonic Distortion: Less than $1 \%$. (Typical $0 \cdot 1 \%$ ) at all output powers and frequencies in the audio band ( 28 V ). Load Impedance: 3 to 15 ohms. Input Im. pedance: 250 Kohms nominal. Power Gain: pedance: 250 Kohms nominal. Power Gain:
$90 \mathrm{~dB}(1,000,000,000$ times) after feedback. 90 dB ( $1,000,000,000$ times) after feedback.
Supply Voltage: 6 to 28 V . Quiescent curSupply Voltage: 6 to 28 V . Quiescent cur-
rent: 8 mA at 28 V . Size: $22 \times 45 \times 28 \mathrm{~mm}$ including pins and heat sink.
Manual available separately 15 p post free.
With FREE printed circuit board and 40 page manual.
$£ 2.98$ Post free

SPECIFICATIONS
Input sensitivities: Radio-up to 3 mV Mag. p.u 3 mV correct to RI.A.A. curve $\pm 1 \mathrm{~dB} .20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u. - up to 3 mV : Aux $-u p$ to 3 mV . Output: 250 mV Signal to noise ratio: better than 70 dB . Channel matching: within 1 dB .
Tone controls: TREBLE +12 to -12 dB at 10 KHz : BASS +12 to -12 dB at 100 Hz . Front panel: brushed alumtnium with black knobs and controls.
Size : $66 \times 40 \times 207 \mathrm{~mm}$.


## Project 605

The easy way to buy and build
 Project 60

Project 605 is one pack containing: one PZ5 iwo 230 s . one Stereo 60 and one Masterlink This new module contains all the input sockets and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting
Complete Project 605 pack with $£ 29.95$
Everything you need to assemble a superb 30 watt high fidelity stereo amplifier without having to solder.

## the world's most advanced high fidelity modules

## Z.30 \& Z.50 power amplifiers <br> Bult, tested and guaranteed with circuits and instructions manual. $2.30 £ 4.48 \quad 2.50 £ 5.48$

The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at 15 w ( $8 \Omega$ ) and all lower outputs. Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Their performance and design are such, however, that $Z .50$ s and $Z .30$ may be used in a far wider range of applications.
SPECFICATIONS (Z.50 units are interchangable with Z.30s in all applications).—Power Outputs: 2.3015 watts R.M.S. into 8 ohms using 35 volts : 20 watts R.M.S. into 3 ohms using 30 volts.
2.5040 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.

Frequency response: 30 to $300.000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. Distortion: $0.02 \%$ into 8 ohms. Signal to noise ratio: better than 70 dB unweighted. Input sensitivity: 250 mV into 100 Kohms (for 15 w into $8 \Omega$ ). For speakers from 3 to 15 ohms impedance. Size: $14 \times 80 \times 57 \mathrm{~mm}$

## Project 60 Stereo F.M. Tuner



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning. printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.
SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range $: 87.5$ to 108 MHz . Sensitivity: $7 \mu \mathrm{~V}$ for lock-in over full deviation. Squelch level: Typically $20 \mu \mathrm{~V}$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. maximum Operating voltage: $25-30 V D C$. Indicators: Stereo on; tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.

## A.F.U. High \& Low Pass Filter Unit

Built tested and guaranteed.
$£ 5.98$
For use between Stereo 60 unit and two 2.30 s or $Z .50$ s. The unit is very easily mounted and is unique in that the cut-off frequencies are continuously variable. As attenuation in the rejected band is rapid ( $12 \mathrm{~dB} /$ octave). there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. There are two filter sections - rumble (high pass) and scratch (low pass). H.F. cut-off ( -3 dB ) variable from 28 KHz to 5 KHz . L.F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at 1 KHz ( 35 V . supply) $0.02 \%$ at rated output. Operating voltage from 15 to 35 V . Current 3 mA . Size: $66 \times 40 \times 90 \mathrm{~mm}$.


## Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ. 6 or PZ.8 where a stabilised supply is essential.

PZ. 530 volts unstabilised $£ 4,98$ P2. 635 volts stabilsed f7.98 PZ.6 35 volts stabilised PZ. 845 volts stabilised (less mains transformer) PZ. 8 mains transformer
f7.98 $\mathbf{£ 5 . 9 8}$

Typical Project 60 applications

| System | The Units to use | togetherwith | Units cost |
| :---: | :---: | :---: | :---: |
| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control, etc. | £4.48 |
| Mains powered record player | Z.30, PZ.5 | Crystal or ceramic P.U. volume control, etc. | £9.45 |
| 12W. RMS continuous sine wave stereo amp. for average needs | $\begin{aligned} & 2 \times 2.30 \mathrm{~s}, \text { Stereo } \\ & 60 ; \text { PZ. } 5 \end{aligned}$ | Crystal, ceramic or mag. P.U.. F.M. Tuner, etc. | £23.90 |
| 25 W . RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers | $\begin{aligned} & 2 \times 2.30 \mathrm{~s}, \text { Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc. | £26.90 |
| 80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. ( 60 W . RMS into 8 ohms) | $2 \times 2.50 \mathrm{~s}$, Stereo 60:PZ.8, mains transformer | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers, etc., controls | £19.43 |
| F.M. Stereo Tuner (£25) \& A.F.U. (£5.98) may be added as required. |  |  |  |
|  |  |  |  |

## Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionics Ltd.. vou are dissatialied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantes in co-operation with Sinclair Radionice Lid.
Each Project 60 modula is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise charge to you, if it is selurned within wo years from the date of purchase. Outside this period of quarentee a mall charge (typically $\mathrm{f1} 00$ ) will be made No charge is made for postage by surface mal. Air Mail is charged at cost.

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ALL OUR DETECTORS ARE BACKED BY A 12 MONTH GUARANTEE


| BRAND NEW GUARANTEED TRANSISTOR |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2G304 |  | 2 N | 20 |  | 20p |  |  | NKT221 |  |
| 2G271 | $10 p$ | 2 N 3134 | 30p | ASY76 | 20 p | BCZ |  | $\bigcirc{ }^{\circ} 28$ |  |
| 2 N 697 | 12 p | 2 N 3135 | 20p | ASZ20 | 30 p |  |  |  |  |
| $2 N 744$ | $17 p$ | 2 N 3702 | 10 p |  | 10p | BF200 | 30p |  |  |
| 2 N 1303 | 15p | 2 N 3707 | 12p | BCl08 | 10 p | BFY5I | $16 p$ |  |  |
| 2 N 1305 | 20p | 2 N 4036 | 55p | BC109 | 10 p |  | dop |  |  |
| 2 N 1309 | 25p | 25002 | 50 p |  | $10 p$ | BFX87 | 25 p |  |  |
| 2 N 1596 | 51 | 25323 | 60p | BC187 | 27p |  | 20 p |  |  |
| 2 N 1613 | 15p | 2 S 324 | 41.20 | BCY30 | 350 | BP $\times 25$ | 30 p | V30/30 |  |
| 2 N 1711 | 24p | AC128 | $18 p$ | $\mathrm{BCY}^{\text {c }}$ | 45p | GET525 | - |  |  |
| 2N1997 | 20p | AD149 | 30 p | BCY34 | 35p | GET875 |  |  |  |
| 2N2270 | ${ }^{60}$ | ASY29 | 20p | 8CY43 | 15p | GET882 |  |  |  |
| SPECIALS |  |  |  |  |  |  |  |  |  |
| ISKRA RESISTORS $\ddagger$ W carbon film low noise all values I $p$ each <br> ON/OFFTOGGLE SWITCH 10p <br> GRELCO 5-WAY SCREW CONNECTOR 13p <br> 2 POLE SW STANDARD JACK SOCKETS 9p <br> KNOBS. lin dia. screw fix. Grey. <br> black 9p <br> RED PUSH BUTTON SWITCH SP. <br> ROUND hole fix 100 |  |  |  |  | SUB MIN ELECTROLYTIC CAPA. <br> CITORS. 5 p each <br> Values $\mu$ F/V; 6.4/25: 10/16: 20164; <br> 80/25: $125 / 10 ; 125 / 16: 25 / 6 \cdot 4 ; 1 / 10$ <br> POLYSTYRENE CAPACITORS 3p each <br> 0.033: 0.047: 0.1 <br> LOUOSPEAKERS. $2 \mathrm{tin} 8 \mathrm{ohm}, 45 \mathrm{p}$ <br> BATTERY CLIPS. PP 3 with leads, $6 p$, <br> ${ }^{\text {ppI }}$, 7p pair |  |  |  |  |

## MARSHALL'S INTEGRATED CIRCUITS <br> NEW LOW PRICES LARGEST RANGE BRAND NEW FULLY GUARANTEED



## D．C．PANEL METERS

TRANSISTORS
$\because 6301 \quad 15 \mathrm{p}-2 \times 293$ $2 G 302$
$2 G 303$
$2 G 306$ 2G306
2G309 $2 G 309$
20344.
293451 2G34513
$2 G 371$ 2 G 37
$2 \mathrm{G37}$
$2 \mathrm{G3} 31$
$2 \mathrm{G4} 4$ 2G3
2G4
2N1 2G417
2N109
2N

2

| 2 N |
| :--- |
| 2 N | 2N388



2N491 $2 N 697$
$2 N 698$ 2N699
2N706 ぼき！
 2N914 2N 2 N 2N930 2N1091 2N113： $2 N 1184$
？N1302 2 N 1304
2 N 1305 2N1300
2N1306
2N1307 2N130
2N1308
2N 1309 $2 N 1309$
$2 N 1483$ $2 N 1507$

$2 N 1613$ | 2N1631 |
| :--- |
| 2N 1637 |

 2N 1701
2N 1702
$2 N 1711$
2 N 1893
$\qquad$
$\qquad$
2N2192
2N2192，
2N2193
－N2193A
2N2194．A
2N2105
2N21904
2N2918A
2N2919
2 N 2219 A
2 S 2220
2N2221

| $2 N 2221 \mathrm{~A}$ |
| :--- |
| 2 N 22224 |
| $\mathbf{N} 2.220 \mathrm{~A}$ |

$2 \mathrm{~N}_{2} 2292 \mathrm{~A}$
2 N 2368
${ }^{2}+2889$
2 N 236 g
2
$2 N 2646$
N271 $21-20$ 2N4303
2N2713 12p $2 \mathrm{Ni}+91$
2N2712
2 N 2713
2 N 2714

$\because \mathrm{N} 2904 \mathrm{~A}$
2 N 290 B
－N2905 $\quad 28 \mathrm{p}$ on $2 \mathrm{O}+919$
$\begin{array}{ll}\mathrm{N} 290 \mathrm{BA} & 26 \mathrm{p} \text { 2N4920 } \\ 28 \mathrm{~N}\end{array}$
28． 28 D 49 O ।
2N290 2 i．
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Practical Electronics October 1972


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| 1 pole | 40p | 40 p | 40p | 40 p | 40 p | 10 p | 40 p | 40 p | 409 |
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| 7 poles | 70 p | 700 | $70 p$ | 950 | 81.20 | 81.20 | 81.20 | 21.95 | 21.95 |
| 8 poles | 70 p | 70p | 70p | 95p | \$1.20 | 21.20 | 81.20 | 28-20 | 42.20 |
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| Voltage | Capacitance | Price | Voltage | Capacitance | Price |
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