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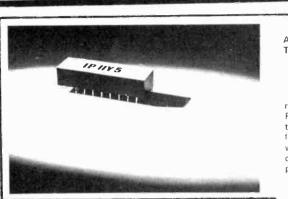
From electrical, radio or car accessory shops or from Antex Ltd., Freepost (no stamp required) Plymouth. PL1 1BR Telephone 0752 67377/8

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The Antex 16 page colour catalogue including details of the X25 ... X25 soldering irons at £1.75 (cheque, P/O, Giro enclosed)

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HY40 IS POWER AMP PERFECTION

Lets face it - an immediate success, the HY40 is here to stay.

HY40 means Hybrid Power, power neatly locked away inside an Intregrated Circuit. Power the modern way, simply mount

only five additional components on a printed circuit board (all of which are supplied with the HY40). Power not only for Hi-Fi, power for Groups, for public address, for industry, power for all.

HY40 is HI-FI POWER ILP are POWER PROUD

In addition to the P.C. board and manual supplied with the HY40 we now include the five remaining components, at minimal cost, needed to complete the assembly of a High Performance Power Amplifier.

By merely combining two HY40s with a Stereo Preamplifier (2 x HY5) and simple Power Supply (PSU45), premium quality stereo may be obtained for a very modest outlay.

The free manual supplied with the HY40 gives clear, easy build instructions for Power Supply; volume, bass, treble and balance controls, together with inputs for Ceramic and Magnetic Pick-ups, Tape, Tuner and Auxiliary functions.

Internally the HY40 is based on conventional and proven circuit techniques developed over recent years.



OUTPUT POWER British Rating 40 WATTS PEAK, 20 watts RMS continuous.

LOAD IMPEDANCE 4–16 ohms INPUT IMPEDANCE 22Kohms at 1Khz.

INPUT SENSITIVITY 300 mV for maximum output.

VOLTAGE GAIN 30db at 1KHz. FREQUENCY RESPONSE 5Hz-60KHz + 1db.

TOTAL DISTORTION less than 1% (typical 0.1%) at all output powers.

SUPPLY VOLTAGE + 22.5 volts D.C.

SUPPLY CURRENT 0.8 amps maximum. PRICE: including comprehensive

manual, P.C. Board and FIVE EXTRA COMPONENTS: MONO £4-40 STEREO £8-80 all post free.

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The HY5 is a unique and revolutionary concept in High-Fidelity pre-amplifiers. Thanks to the latest techniques, all feedback and equalization networks are, for the first time, combined into an integrated pre-amplifier circuit.

Simply by adding volume, treble, bass potentiometers and only three stabilizing capacitors, which are supplied, your HY5 is complete and ready for use.

The HY5 provides equalization for almost every conceivable input. This years developments in equalization technique enables precise correction for both output voltage and frequency response for any crystal or ceramic cartridge. Yet another feature of the HY5 is its inbuilt stabilization circuit, allowing it to be run off any unregulated power amplifier supply.

The HY5 contains a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo preamplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY40 and PSU45 forms a completely integrated system.

INPUTS

Magnetic Pick-up (within ±1db RIAA curve) 2mV. Tape Replay (external components to suit head). 4mV. Microphone (flat) 10mV. Ceramic Pick-up (equalized and compensatable) 20 – 2000mV variable. Tuner (flat) 250mV. Auxiliary 1 250mV. Auxiliary 2 2–20mV. OUTPUTS Main Pre-amp output 500mV. Direct tape output 120mV. ACTIVE TONE CONTROLS Treble ±12db. Bass ±12db. INTERNAL STABILIZATION

INTERNAL STABILIZATION Enables the HY5 to share an unregulated supply with the Power Amplifier. SUPPLY VOLTAGE 15–25 volt. SUPPLY CURRENT 5mA approx. OVERLOAD CAPABILITY better than 28db on most sensitive input infinite on tuner and auxi. OUTPUT NOISE VOLTAGE 0.5mV. PRICE

Mono £3-60 Stereo £7-20

POWER SUPPLY PSU45



The PSU45 is specifically designed to supply, simultaneously, your HY40 (in mono or stereo format) and one or two HY5s.

Spec.

PSU45 <u>+</u> 22.5 volts, 2 amps simultaneously.

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250V 0.068µF 6p 100V 22µF 9p	Voltage 100V 100V 100V 100V 100V 100V 100V 250V 250V 250V 250V	$\begin{array}{c} \textbf{Capacitance} \\ 0^{-1}\mu\text{F} \\ 0^{-1}5\mu\text{F} \\ 0^{-2}2\mu\text{F} \\ 0^{-3}3\mu\text{F} \\ 0^{-4}7\mu\text{F} \\ 0^{-4}7\mu\text{F} \\ 0^{-6}8\mu\text{F} \\ 0^{-0}15\mu\text{F} \\ 0^{-0}15\mu\text{F} \\ 0^{-0}22\mu\text{F} \\ 0^{-0}23\mu\text{F} \\ \end{array}$	6р 6р 9р 10р 15р 5р 5р 5р	10V 10V 25V 25V 25V 25V 25V 25V 25V 25V 35V 35V	22µF 470µF 10µF 100µF 220µF 100µF 2,200µF 2,200µF 4.7µF 220µF	7p 11p 7p 9p 14p 22p 42p 7p 14p
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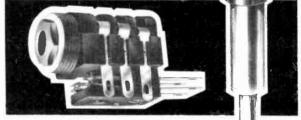
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Practical Electronics July 1972



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The exclusive Duo loudspeaker systems are incomparable for quality within their price range. Large speakers in extremely substantial cabinets. There's a choice of the Duo II's for the smaller room or the big Duo III's for real bass response.

Check through the technical specification for convincing evidence of the true value and excellence of Viscount III suites.

SPECIFICATION. 14 watts per channel into 3-4 ohms (suitable 3-15 ohms). Total distortion at 10W, at 1 kHz, 0.1%, P.U.1 (for ceramic cartridges): 180mV into 3 Mez. P.U.2 (for magnetic cartridges). 4 mV at 1 kHz into 47 K estimated within ± 1 dB R.I.A.A. Radio: 150 mV into 220K. (Sensitivities given at 10 power). Tape out facilities; headphone socket, power out 250 mV per characteristics. Bass: +12 dB to -17 dB at 60 Hz at 8 s filter: 6 dB per octave cut. Treble control: treble +12 dB to -0 Hz at 5 kHz. Treble filter: 12 dB per octave. Signal to noise ratio: (all controls at max) P.U.1 and radio -65 dB. P.U.2: -58 dB. Crosstalk better than -55 dB on all inputs. Overload characteristics better than 26 dB on all inputs.



SPEAKERS Duo Type II. Size: Approx. 17in × 104in × 64in. Drive unit: $13in \times 8in$ with parasitic tweeter. Max. power 10 watts, 8 ohms. Simulated teak cabinet. £14 pair+£2 p&p.

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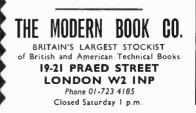
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$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Pak No. fp Q_2 16 White U1 120 Glass sub-min. general purpose germanium diodes 0.50 Q_4 6 White U2 60 Mixed germanium transistors AF/RF 0.50 Q_4 6 Match U2 60 Mixed germanium transistors AF/RF 0.50 Q_4 6 Match U3 75 Germanium gold bonded diodes sim. 0.55, 0.447 0.50 Q_4 4 0.C72 U4 40 Germanium transistors like OC81, AC128 0.50 Q_8 4 AC129 U4 40 Germanium transistors NPN sim. BS V95A, 2N706 0.50 Q_1 2 AF11 U5 60 200nA sub-min. Sil. diodes 0.500 Q_1 2 AF11 U6 30 Silicon planar transistors NPN sim. BS V95A, 2N706 0.50 Q_1 2 AF11 U8 50 Sil. planar diodes 250m A, $OA/200/202$ 0.500 Q_1 2 AF11 U3 20 Mixel volts 1 watt Zener.diodes 0.500 Q_1 2 GETE U1 20 Nixel volts 1 watt Zener.diodes 0.500 Q_1 3 Match U3 25 NPS silicon planar transistors TO-	n_{00} trans. PNP AF 0.50 n_{00} trans. FNP 0.50 $type$ trans. OC44/45781/81D 0.50 $transistors$ 0.50 $type$ trans. 0.50 $type$ trans. 0.50 $type$ trans. 0.50 7 type trans. 0.50 26 sile poxy trans. 0.50 1 ST14 & 2 ST140 0.50 1 ST14 & 2 ST140 0.50 1 ST14 & 2 ST140 0.50 1 ST141 & 2 ST140 0.50 1 ST14 & 2 ST140 0.50 1 ST141 & 2 ST140 0.50 1 ST16 A 1.20 0.50 1 ST16 A 1.40 0.50 1 ST16 A 1.50 0.50 3 S MAT 101 & MAT 121 0.50
TRIACS VBOM 2A 6A 10A TO-1 TO-66 TO-88 ½ ½ ½ ½ 100 30 50 76 200 50 60 90 400 70 75 1-10 DIACS FOR USE WITH	FULL RANGE OF ZENER DIODES VOLTAGE RANGE 2-33V. 400.nlv (DO-T Gase) Jap ca. 14W (Top- Hat) 18p ca. 16W (SO-10 Stud) 25p ca. All Mild tested 5% (O- and marked. State voltage required.	$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
TRIACS String String<	BECTIFIERS on heat sink. 100 PIV. 90p each. JUMBO COMPONENT PAKS MIXED ELECTRONIC COMPONENTS Exceptionally good value	124 20 0 cermanium P Amp restincts (0.3) in 0.0 304 11 0.50 \$\frac{3}{4}\$ 7, 831, N\$ 125 25 300 Mc/s N N's Millicon transitions 2N708, BNY77 0.50 \$\frac{3}{3}\$ 3, 831. 126 30 Faat switching silicon divers 2N708, BNY77 0.50 \$\frac{3}{3}\$ 3, 831. 129 10 1-Amp SCR's TO-5 can up to 600 PTV CR31/25-600 1.00 \$\frac{3}{3}\$ 8, 77. 120 10 1-Amp SCR's TO-5 can up to 600 PTV CR31/25-600 1.00 \$\frac{3}{3}\$ 8, 77. 123 20 Ril. Planar NPN trans. low noise amixed volts, 3-18 0.50 \$\frac{3}{3}\$ 7 NPN 123 25 Plastic case 1 amp allicon tre-tifters NM000 series 0.50 \$\frac{4}{4}\$ 7 NPN 123 30 Sil. PNP alloy trans. TO-5 BCY26, 28302/4 0.50 \$\frac{4}{4}\$ 7 BC10 123 25 Plastic case 1 amp allicon tre-tifters NM000 series 0.50 \$\frac{4}{4}\$ 7 NPN 123 25 Plastic case 1 amp allicon tre-tifters NM000 series 0.50 \$\frac{4}{4}\$ 7 NPN 123 25 Sil, planar trans. PNP TO-18 2N2906 0.50 \$\frac{4}{4}\$ 7 RNN	33 N PN sil. trains. 0-50 trans. 4 × 253703, 3× 293702 0-50 trans. 4 × 253704, 3× 293705 0-50 anip. 4 × 293707, 3× 293706 0-50 trans. 205172 0-50 trans. 4 × 20106, 3× 20500 0-50 trans. 4 × 20106, 3× 1000 0-50 trans. 4 × 100106, 3× 1000 0-50
And coded TO-18 case. FREE One 50p Pak of your own choice free with orders valued £4 or over. SPECIAL OFFER 2N2926 (Y) (O), 10 for 50p, 23 for £1, 29,000 TO CLEAR.	Resistors, ccapacitors, pots, electrolytics and coils plus many other useful items. Approxi- mately 3lbs in weight. Price incl. P. & tr. £1.50 only Plus your salisfaction or money back sustante. BRAND NEW TEXAS GERM. TEANSISTORS	137 30 8il. alloy trans. 80-2 PNP, 0C200 28322 0.50 647 is NPN 138 20 Fast switching sil. trans. NPN, 400Mc/s 2N3011 0.50 648 16 WPN 138 20 Fast switching sil. trans. NPN, 400Mc/s 2N3011 0.50 649 4 NPN 139 30 RF germ. PNP trans. 2N1303/5 T0-5 0.50 650 718 SY 140 10 Dual trans. 6 lead T0-5 2N2060 0.50 652 8 BY10 141 25 RF germ. trans. T0-1 OC45 NKT72 0.50 652 25 Sil. 121 10 VifP germ. PNP trans. T0-1 NKT667 AF117. 0.50 73 25 Sil. 123 25 Sil. trans. plastic T0-18 A.F. BC113/114 0.50 PRINTED PRINTED	5 NYN TO-5 trans. 0-50 bich gain 32 kBC167, 35 RC188 0-60 70 NPN trans. TO-18 0-50 70 NPN trans. TO-18 0-50 928 NPN awitch TO-18 0-50 0-50 93 NPN trans. 300MH2 0-50 93 NPN trans. 0-50 0-50 93 NPN trans. 1-00 0-50 94 new 1-50 1-50 CIRCUITSEX-COMPUTER
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Order No.		100 սր £թ	Order No.	£p	£p.	400 up ≴p
$BP00 = 8N^{2}400$	£μ £μ 0·15 0·14	±p 0·12	BP86 = SN7486	0.32	0.30	0.28
BP00 = SN 400 BP01 = SN 401	0.15 0.14	0.12	BP90 = SN7490	0-67	0.64	0.58
BP01 = SN7401 BP02 = SN7402	0.15 0.14	0.12	BP91 = 8N7491AN	0.87	0.84	0.78
BP02 = SN7402 BP03 = SN7403	0.15 0.14	0.12	BP92 = 8N7492	0.87	0.64	0.58
BP03 = 8N7403 BP04 = 8N7404	0.15 0.14	0.12	BP93 = SN7493	0.67	0.64	0.58
BP04 = 8N7404 BP05 = 8N7405	0.15 0.14	0-12	BP94 = SN7494	0.77	0.04	0.68
BP03 = SN7403 BP07 = SN7407	0.18 0.17	0.12	BP95 = 8N7495	0.77	0.74	0.68
BP07 = SN7407 BP08 = SN7408	0.18 0.17	0.16	BP96 = SN7496	0.77	0-74	0.68
BP08 = 8N7408 BP09 = 8N7409	0.18 0.17	0.16	BP100 = 8N74100	1.75	1.65	1.55
BP10 = 8N7410 BP10 = 8N7410	0.15 0.14	0.12	BP100 = SN74100 BP104 = SN74104	0.97	0-94	0-88
BP10 = SN7410 BP13 = SN7413	0.29 0.26	0.24	BP105 = 8N74105	0-97	0-94	0.88
BP16 = 8N7416	0.43 0.40	0.38	BP107 = SN74107 BP107 = SN74107	0.40	0-38	0-36
BP17 = SN7417 BP17 = SN7417	0.43 0.40	0.38	BP110 = SN74110	0-55	0-53	0.50
BP17 = SN7417 BP20 = SN7420	0.15 0.14	0.12	BP111 = 8N74111	1.25	1.15	1.00
BP30 = SN7420 BP30 = SN7430	0.15 0.14	0.12	BP118=8N74118	1.00	0.95	0.90
BP40 = SN7440	0.15 0.14	0.12	BP(19 = SN74119)	1.35	1.25	1.10
BP41 = 8N7441	0.67 0.64	0.12	BP121=SN74121	0.67	0.64	0.58
BP42 = SN7442	0.67 0.64	0.58	BP141 SN74141	0.67	0.64	0.58
BP42 = SN7442 BP43 = SN7443	1.95 1.85	1.75	BP145 = SN74145	1.50	1.40	1.30
BP43 = SN7443 BP44 = SN7444	1.95 1.85	1.75	BP150 = 8N74150	1.80	1.70	1.60
BP45 = SN7445	1.95 1.85	1.75	BP151 = SN74151	1.00	0.95	0.90
BP46 = 8N7446	0.97 0.94	0-88	BP153 = 8N74153	1.20	1.10	0.95
BP47 = SN7447	0.97 0.94	0.88	BP154 = 8Ne74154	1.80	1.70	1.60
BP48 = 8N7448	0.97 0.94	0.88	BP155 = 8N74155	1.40	1.30	1.20
BP50 = 8N7450	0.15 0.14	0.12	BP156 = SN74156	1.40	1.80	1.20
BP51 = SN7451	0.15 0.14	0.12	BP160 = SN74160	1.80	1.70	1.60
BP53 = 8N7453	0.15 0.14	0.12	BP161 = SN74161	1.80	1.70	1.60
BP54 = 8N7454	0.15 0.14	0.12	BP164=8N74164	2.00	1.90	1.80
BP60 = 8N7460	0-15 0-14	0.12	BP165 = SN74165	2.00	1.90	1.80
BP70 = SN7470	0.29 0.26	0.24	BP181=SN74181	2.75	2.80	2.40
BP72 = 8N7472	0.29 0.26	0.24	BP182 = SN74182	0.97	0.94	0.88
BP73=8N7473	0.37 0.35	0.32	BP190=8N74190	3.50	3-25	3-00
BP74 = 8N7474	0.37 0.35	0.32	BP191=8N74191	3.50	3.25	3.00
BP75=8N7475	0.47 0.45	0.42	BP192=8N74192	2.10	1.95	1.75
BP76 = 8N7476	0.43 0.40	0-38	BP193=8N74193	2.10	1.95	1.75
BP80 = 8N7480	0.67 0.64	0.58	BP125 = SN74195	1.10	1.05	0-95
			BP196 = SN74196	1.80	1.70	1.60
BP81=SN7481	0.97 0.94	0-88	BP197 = 8N74197	1.80	1.70	1-60
BP82 = SN7482	0-97 0-94	0.88	BP198=8N74198	5.50	5-00	4.00
BP83=SN7483	1.10 1.05	0.95	BP199=SN74199	5.50	5-00	4-00

PRICE-MIX. Devices may be mixed to qualify for quantity prices.

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Owing to the ever increasing range of TTL 74 Series, please check with us for supplies of any devices not listed above, as it is probably now in stock. WARE 3442.

LINEAR I.C's	FUI	LL SP	EC.	ROCK					
		Price	3	LOGIC Type	DTL	. 93	0 Se	Price	.C's
Type No.	1-24	22-99	100 up	No.			1 - 24	25-99	100 up
BP 201C-8L201C	63 p	58p	45p	BP930			12p	11p	10p
BP 701C -8L701C	63p	50p	45p	BP932			13p	12p	11p
BP 702C- SL702C	63 p	50p	45p	BP933 BP935			13p 13p	12p 12p	11p 11p
BP 702 - 72702	53p	45p	40p	BP935 BP936			13p	120	11p
BP 709 72709	53p	45p	40p	BP944			13p	12p	lip
				BP945			25p	24 p	22p
BP 709P -μA709C	53p	45p	40p	BP946			12p	11p	10p
BP 710-72710	53p	45p	40p	BP948			25p	24p	22p
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BP 74172741	75p	60p	50p	BP962			12p	11p	10p
μA 703C-μA703C	43 p	35p	27p	BP9093			40p	38p 38p	35p 35p
				BP9094			40p 40p	38p	35p
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TAA 293-	90p	75p	70p	BP9099			40p	38p	35p
TAA 350	170p	158p	150p	Devices ma tity, price.	Larg		ro qu juanti	alify fo	ces or
S.G.S. EA 1000 £2.6				anolication				ies.onl	

NUMERICAL INDICATOR TUBES

CD66	GRI16	3015F	
170 min	175 min	ō trin	
2.3	14	8	All indicators
16	13	9	point: All side
47	32	22	data for all
19	13	12 wide	types available on request.
BP41 or 141	BP41 or 141	BP47	
£1.70	£1.55	£1·90	1
	170 min 2·3 16 47 19 BP41 or 141	170 min 175 min 2·3 14 16 13 47 32 19 13 BP41 or 141 BP41 or 141	170 min 175 min 5 min 2·3 14 8 16 13 9 47 32 22 19 13 12 wide BP41 or 141 BP41 or 141 BP47



- ★ Frequency Response 15Hz to 100,000--1dB.
- ★ Load-3, 4, 8 or 16 ohms.
- \pm Distortion—better than 0.1% at
- lkHz.

 \pm Signal to noise ratio 80dB.

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£3·25p each

★ Supply voltage 10-35 volts

★ Overall size 63mm×105mm×13mm

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Manufacturers' "Fall outs"—out of spec. devices including functional units and part function but classed as out of spec. from the manufacturers' very rigid specifications. Tieleal for learning about LC's and experimental work.

them in the featuring more	
Pak No. U1C930 = $12 \times \mu A 930$ U1C932 = $12 \times \mu A 932$ 11C933 = $12 \times \mu A 933$ U1C935 = $12 \times \mu A 933$ 11C935 = $12 \times \mu A 933$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
$\begin{array}{l} {\rm UIC936} = 12 \times \mu A \ 936 \\ {\rm UIC944} = 12 \times \mu A \ 944 \\ {\rm UIC945} = 8 \times \mu A \ 945 \\ {\rm UIC946} = 12 \times \mu A \ 946 \\ {\rm UIC946} = 12 \times \mu A \ 946 \\ {\rm He} \end{array}$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
Data Booklet available for the U1000 = 12×7400 M 50p U1001 = 12×7401 M 50p U1002 = 12×7401 M 50p U1002 = 12×7402 M 50p	he BP930 Series, PRICE 13p. UIC46 = 5 × 7446N 50p UIC81 = 5 × 7481N 50p UIC47 = 5 × 7447N 50p UIC82 = 5 × 7482N 50p UIC88 = 5 × 7448N 50p UIC83 = 5 × 7483N 50p
$\begin{array}{l} 1^{+}IC03 = 12 \times 7303N & 50p \\ 1^{+}IC04 = 12 \times 7404N & 50p \\ 1^{+}IC05 = 12 \times 7405N & 50p \\ 1^{+}IC05 = 12 \times 7410N & 50p \\ 1^{+}IC13 = 12 \times 7410N &$	$\begin{array}{llllllllllllllllllllllllllllllllllll$
$\begin{array}{l} U1C20 = 12 \times 7420N \ \ 50p \\ U1C40 = 12 \times 7440N \ \ 50p \\ U1C41 = 5 \times 7441AN50p \\ U1C42 = 5 \times 7442N \ \ 50p \\ U1C43 = 5 \times 7443N \ \ 50p \\ U1C44 = 5 \times 7443N \ \ 50p \\ U1C44 = 5 \times 7443N \ \ 50p \\ U1C44 = 5 \times 7444N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \\ U1C44 = 5 \times 744N \ \ 50p \ \ \ \ 50p \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \$	$\begin{array}{llllllllllllllllllllllllllllllllllll$
$U1C45 = 5 \times 7445N$ 50p	$U1080 = 5 \times 7480N 80p$





LOW COST BRIDGE RECTIFIERS

DCoutput		Type Nos.
Amps	Volts	
10 when mounted	60 125 250 375	PM7A1 PM7A2 PM7A4 PM7A6
15 when mounted	60 125 250 375	PM7A10 PM7A20 PM7A40 PM7A60

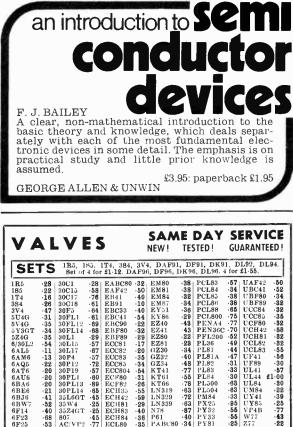
AMP tags electrically isolated from mounting bracket. Mount them on a chassis, the equipment box, transformer housing etc.



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	6AM6	·13	30P4	.57	ECC83	·35	GZ32	-40	PL81A	-47	UF41	-56
	6AQ5	.22	30P12	.72	ECC85	.34	GZ34	-48	PL82	.31	UF89	-30
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	6BA6	-20	30PL13	-89	ECF82	-26	KT66	-78	PL500	-63	UL84	-80
	6BE6	.21	30PL14	-65	ECH35	.55	LN319	·63	PL504	-63	UM84	.22
	6BJ6	-41	35L6GT	.45	ECH42	-59	LN329	.72	PM84	-33	UY41	·39
	6BW7	-52	35 W 4	·25	ECH81	.29	LN339	·63	PX25	-95	UY85	·25
	6F14	-40	35Z4GT	·25	ECH83	.40	N78	-87	PY32	-55	VP4B	-77
	6F23	-68	807	.45	ECH84	·36	P61	.40	PY33	- 55	W77	-43
	6F25	.53	AC/VP2	.77	ECL80	-35	PABC80	-34	PY81	.25	Z77	-22
	6J7G	.24	B349	.65	ECL82	-31	PC86	.47	PY82	·25	Transi	
	6K70	.12	B729	·62	ECL86	-35	PC88	-47	PY83	·28	AC107	-17
	6K8G	.17	CCH35	-67	EF39	-38	PC96	-42	PY88	•33	AC127	·18
	6Q7G	·35	CY31	·30	EF41	-60	PC97	-39	PY800	-84	AD140	-37
	6SN7GT	.30	DAF91	-22	EF80	·23	PC900	·31	PY801	-34	AF115	·20
	6V6G	·28	DAF96	·36	EF85	·28	PCC84	-29	R19	·30	AF116	-20
	6V6GT	-28	DF33	-38	EF86	-30	PCC85	·25	R20	-56	AF117	·20
	6X4	-23	DF91	-16	EF89	-26	PCC88	·40	U25	·64	AF118	-48
	6X5GT	-28	DF96	·36	EF91	.13	PCC89	·45	U26	·56	AF125	-17
	10P13	-58	DH77	·20	EF92	-30	PCC189	·48	U47	·64	AF127	-17
	12AH8£2	.95	DK32	-33	EF98	·65	PCC805	·56	U49	-56	OC26	-25
ŀ	12AT7	-17	DK91	·28	EF183	·28	PCF80	·28	U50	·26	OC44	-12
į.	12AU6	·20	DK92	·50	EF184	·81	PCF82	·33	U52	•31	OC45	-12
	12AU7	·20	DK96	-45	EH90	·35	PCF86	·48	U78	-24	OC71	-12
	12AX7	·22	DL35	-40	EL33	·55	PCF800	·58	U191	-59	OC72	-12
	19BG6G		DL92	·26	EL34	-45	PCF801	·28	U193	•42	OC75	-12
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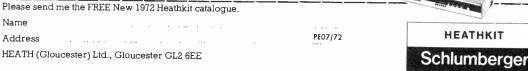
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SYSTEMS ARE GO!

PRIVATE constructors have become accustomed to using integrated circuits, singly or in small quantities, in all manner of circuits. In many instances, though, it could be argued that the use of i.c.s is not really justified—that a few discrete semiconductors could be employed satisfactorily for the same purpose. This will always be a debatable point.

Still it has to be admitted that small projects do not adequately demonstrate the actual or potential capabilities of i.c.s. Their unique features are better exploited when they are used in quantity to make up some extensive and complex system: a system such as would be extravagant in the extreme—if even practicable—were discrete components to be used throughout.

Another expansive era in d.i.y. electronics beckons. Already the imagination and ambition of both designer and constructor have been released from many old inhibitions, imposed by visions of frightening circuit complexity, enormous size of the final assembly, and cost. The sky will be the limit, or nearly so, in the future.

At this juncture it is right to pause and recognise that many constructors are quite happy to indulge in their hobby just to the extent of the more popular and relatively modest projects, whether based on discrete or integrated devices. Such needs will always receive our careful attention and be properly catered for in P.E.

On the other hand, we are sure that amongst our readers there are quite a number who would welcome the opportunity to tackle a project of somewhat larger proportions, from time to time. We hope therefore to present on occasion some special advanced design that offers the constructor a chance to take full advantage of i.c.s and build an item of equipment of a more ambitious nature than hitherto. As a start, we introduce this month the P.E. Digi-Cal—the first high speed calculator designed specially for the constructor.

Small desk calculators are a major growth area in the electronics industry. In this field we witness some most remarkable scaling-down in physical size, due to the adoption of large scale integration techniques. At present l.s.i. is exclusively for equipment manufacturers. Thus the home constructor cannot as yet (if ever) expect to match size for size the smaller of these commercial calculators. This is of course a basic fact that has to be faced in other branches of electronics also.

Yet, when all is considered, this is no very serious limitation. There are countless needs that do not demand an instrument capable of being carried in the vest pocket! Wherever much calculating work is regularly encountered, at home, school, or business, the Digi-Cal will amply pay for the small area of table or desk it occupies. F.E.B.

THIS MONTH

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AT LAST.

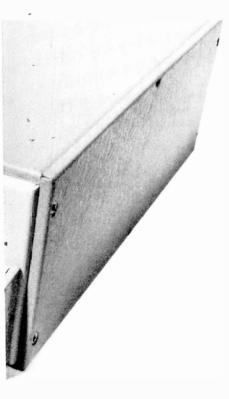
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DESK-TOP

CALCULATOR FOR CONSTRUCTORS

BY R.W.COLES



PE

We have all become familiar with the host of mechanically operated calculating machines, typified by the supermarket cash-register, which have been with us for many years. The ingenious mechanisms used in these machines can even be coaxed to perform multiplication or division in a simple way, but they are not up to the standard required for general mathematical problems, and have never made much of an inroad into this sphere, where the slide-rule has reigned supreme until very recently.

With the advent of the digital computer employing transistors, and later, integrated circuits, it was soon recognised that it would be possible to build small computers on the lines of mechanical calculators, which would give much simpler operation and a more versatile problem solving capability.

We are now witnessing the heyday of the offspring of the computer/ mechanical calculator marriage, and the numbers of small electronic problem solvers becoming available is increasing dramatically every year, bringing big-system electronics to many desk corners.

AMATEUR CONSTRUCTORS

Up to now there has not been an electronic calculator design suitable for amateur constructors, despite the availability of all the necessary integrated circuits. This is mainly due to the fact that the know-how accumulated by the manufacturers of commercial calculators is very jealously guarded, and, to the uninitiated, calculator circuitry does seem quite complex.

Digi-Cal sets out to redress this situation, being designed specifically for home construction and simplified as far as possible without sacrifice of performance. Digi-Cal is a fast flexible tool for performing calculations required in the home, school or laboratory. It can be built using basic techniques and requires no access to expensive test equipment such as oscilloscopes. The prototype was built entirely on the kitchen table!

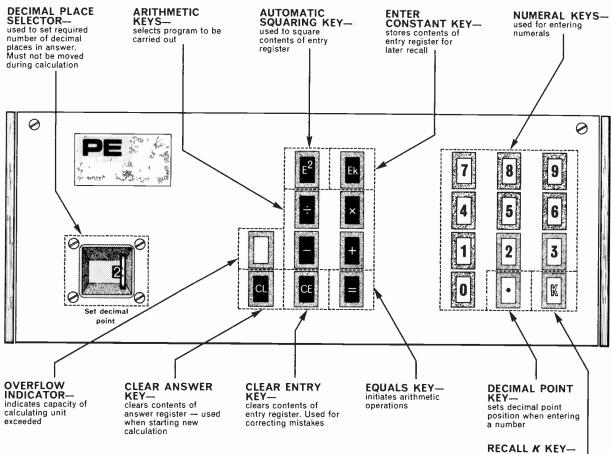


Fig. 1.1. Key functions and layout of the Digi-Cal keyboard

recalls stored constant J to entry register

The basic Digi-Cal design has been intended from the outset to be as open-ended as possible to allow for extra features and constructor options. The arithmetic programmes are all easily altered by simply wiring in extra diodes, the constant store and the auto-squaring facility may be omitted and included at a later date, or even omitted altogether, and there are many possible performance improvements for which the design allows without major reconstruction.

All the logic is built on removable cards for easy access, and the complete calculator is housed in a modified "Contil" Mod-2 case with a matching keyboard, making Digi-Cal easily portable.

USING DIGI-CAL

The manipulation of numbers and the initiation of an arithmetic process is carried out using the keyboard, a vital part of any calculator and one which must be carefully thought out both mechanically and electronically. The keyboard is shown in Fig 1.1.

Entries are made by pressing figure keys in the correct sequence, interposing the decimal point where required if necessary. When the first figure key is pressed the display automatically switches from the last anwer to display the new entry, the old answer remaining in store. As subsequent figure keys are pressed they enter from the right taking the decimal point with them, in its correct position, as soon as it is entered. See Figure Entry. The actual number of figures after the decimal point is immaterial at this point since Digi-Cal positions the numbers automatically before starting a calculation. However, if too many figures are entered, the entry register locks to prevent any further figure inputs, so that if the keyboard is set to give answers to three places, after three places of decimals no further figures can enter the calculator.

With a number entered in the entry register the next step is to select the arithmetic function required, and this is achieved by pressing the appropriate black key. The function entered will remain valid for all subsequent calculations until a different function is required and another key pressed. This means that a whole series of numbers may be added, for example, with only one depression of the add (+) key. Each time a new number is entered the equals key is pressed and the number is automatically added to the previous total.

Pressing the equals key starts the calculation, and when the programme for that function is complete the display switches to the answer register to display the result.

All operations are performed on the previous answer, so when starting with an empty answer register (after it has been cleared by the clear key) the first number entered must be added into it before the calculation proper begins. If a mistake is made when entering figures, the entry register can be cleared with the clear entry (CE) key.

ENTRIES UP TO SIX DIGITS WITH FLOATING DECIMAL POINT CAN BE MADE, ANSWERS ARE PRODUCED UP TO EIGHT DIGITS LONG WITH THE DECIMAL POINT IN ONE OF FOUR PRESELECTED POSITIONS

ENTRIES AND ANSWERS ARE DISPLAYED ON EIGHT SEVEN-SEGMENT INDICATORS WITH AN ADVANCED LEADING-EDGE ZERO-SUPPRESSION SYSTEM

PROBLEMS CAN BE SIMPLE TO USE. ENTERED AS THEY WOULD BE WRITTEN DOWN

AVAILABLE TTL EMPLOYS READILY INTEGRATED CIRCUITS THROUGHOUT

OPERATIONS USING A CONSTANT

When operation with a constant is required the constant is entered in the entry register in the normal way, except that it must be positioned with the decimal point in the proper place, and then the EK key is pressed to store the constant for future use. When the constant is required the white K key is pressed, which instantly recalls the stored number to the entry register ready for use. The constant remains in store until replaced or the machine switched off.

The following examples will show just how simple Digi-Cal is to use and how useful it can be to solve a wide range of problems encountered in all walks of life.

FIGURE	ENTRY
SET DECIMAL PLACES	DISPLAY
PRESS	0.000
2	2.
3	23.
	E 5
4	534
5	2345
+	23.450

EXAMPLE ONE

CHAIN CALCULATION (14-35+2-6+200) 0-58 9.8 SET DECIMAL PLACES 3 DISPLAY PRESS 3 4 5 • = 6 15 0 2 0 216 = 5 8 1258 8 =

EXAMPLE ONE

2

This example shows how a simple arithmetical problem can be easily solved using Digi-Cal.

First the number of decimal places required in the answer is set on the decimal place selector. The first number is then entered by pressing the appropriate keys. The add key is then depressed and then the equals key to enter the first number into the calculator. The next number is then entered and added to the previous one by simply pressing the equals key. Similarly with the third number.

We now wish to multiply this total by 0.58 so the multiply key is pressed, 0.58 is entered and the equals key completes the multiplication. The divide key is then pressed, 9.8 is entered and a final press of the equals key completes the calculation.

EXAMPLE TWO OPERATION USING CONSTANT STORE 1789.50+3% SET DECIMAL PLACES 2 DISPLAY PRESS 8950 9 8 5 8958 0 3

EXAMPLE TWO

This second example illustrates the use of the stored constant. We wish to add 3% of 789.5 to 789.5. The calculation required is thus 789.5 + (0.03 × 789.5). The number 789.5 is used twice, so to facilitate the calculation it is stored so that it will not have to be entered twice.

Of course, since pounds and pence are involved it is natural to set the number of decimal places to 2. The number 789.5 is then entered but the EK key is pressed before the add and equals keys. The multiply key is then depressed and 0.03 is entered. The equals key then gives 3% of 789.5. Pressing the add and the K key followed by the equals key adds the stored number (789.5) to the previous answer to give the final answer.

PRINCIPLES OF OPERATION

Digi-Cal is built entirely of gates and flip-flops, and operates in a similar manner to the large computers which compile gas bills or calculate wage checks. The main difference is that whereas a large business or scientific computer can be programmed and reprogrammed with ease, Digi-Cal and similar calculators are fixed programme machines, the programme being established at the construction stage.

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Another basic difference is that while a large computer performs all its calculations in binary arithmetic, Digi-Cal uses a combination of binary and decimal working known as Binary Coded Decimal arithmetic.

In Binary Coded Decimal (B.C.D.) each decimal digit (e.g. a six or a nine) is represented by a separate group of four binary digits (i.e. 0110 and 1001 respectively). Four binary digits have sixteen possible combinations, but to encode decimal data only the first ten of these are utilised, the other six combinations are redundant and represent invalid data if they should occur.

Representing decimal data using these four bit (bit means binary digit) groups is quite different from their representation using straight binary code as can be seen in the following example:

Decimal	362
Straight binary	101101010
B.C.D.	0011 0110 0010

Binary Coded Decimal operation is used in Digi-Cal because it eliminates a good deal of the conversion circuitry which would be required if straight binary code were to be used for the arithmetic operations, since the input and output data must be in decimal to make it simple for the average user.

B.C.D. ARITHMETIC

Adding in straight binary is very simple and follows the following rules:

0 + 0	= ()		
1 + 0	=	1		
0 + 1				
1 + 1	= (), c	arry	1

With B.C.D. working, a complication arises because although the rules given above do apply when adding individual bits within the group of four digits, as soon as the value of those four digits exceeds nine (not fifteen as in straight binary) a carry has to be generated which is added into the next higher group of four B.C.D. digits. To add 16 and 29 together using B.C.D. we proceed thus:

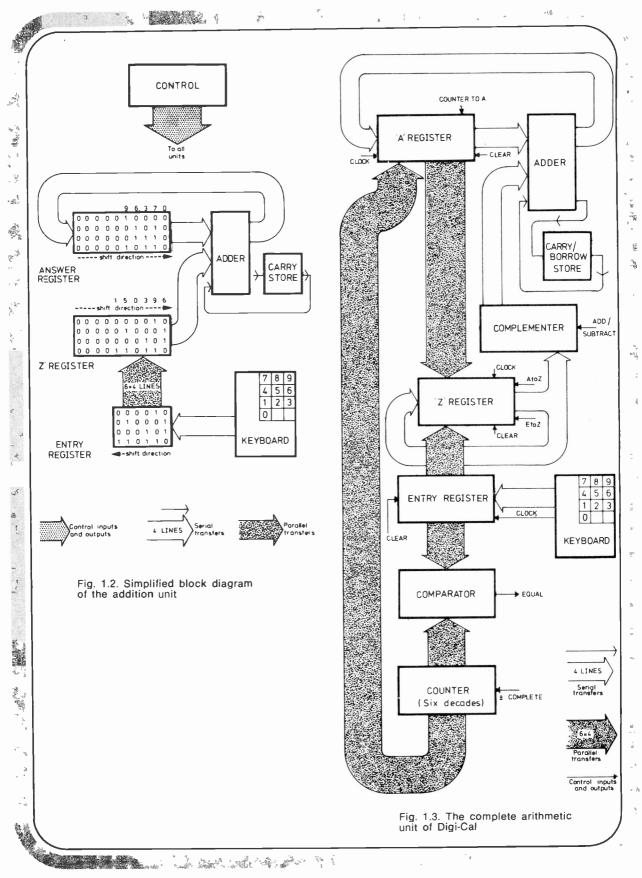
	DECIMAL		B.C.D.
1	6	0001	0110
2	9	0010	1001
1 🐳		1 🔸	
4	5, carry 1	0100	0101, carry 1

In Digi-Cal all arithmetic operations can be broken down to the simple processes described above, and the way that this is achieved in practice can be seen by referring to Fig. 1.2 which is the simplified block diagram of the heart of Digi-Cal.

SYSTEM OPERATION

As each key is pressed the numbers are coded into B.C.D. and then enter the Entry (E) register from the right, one after another. When the required entry has been completed the addition is initiated by pressing the "equals" key which starts the programme sequence.

The Z register is first cleared of any data it may contain and then the entire contents of the E register are transferred in parallel to it $(6 \times 4, 24)$ bits). It is at this stage of the sequence that Digi-Cal appears in Fig. 1.2.



Both numbers are positioned ready for addition, and to implement this the A register and the Z register are clocked ten times with a burst of ten pulses from the control circuits. Each of the ten pulses presents two new B.C.D. numbers to the adder, which produces a sum and stores a carry if necessary.

When the next pulse arrives the previous answer is shifted into the far end of the A register, so that after the complete addition the first answer is stored in the right-hand location of the A register, and the last answer in the left-hand location, as they should be. In a nutshell then, Digi-Cal carries out parallel B.C.D. but serial decimal addition.

OTHER OPERATIONS

Subtraction is carried out in exactly the same way as addition except that the number to be subtracted is converted to its complement (i.e. each digit subtracted from nine) form before the addition takes place.

Multiplication and division are carried out by performing successive additions and subtractions respectively, and several additions to the basic circuit of Fig. 1.2 are necessary to achieve this. A more complete circuit of the arithmetic section of Digi-Cal is shown in Fig. 1.3. As can be seen, a counter, a comparator and a complementer have been added, along with a number of new interconnections.

MULTIPLICATION

To perform a multiplication the multiplicand is transferred to the Z register from the A register which is then cleared. The multiplier (stored in the E register) is compared with the contents of the counter which is connected to count each complete ten digit addition.

Additions are started by supplying batches of ten clock pulses to the A and Z registers, and this continues until the counter has counted up to the same number as is stored in the E register, whereupon the comparator indicates equality and stops the clock.

The contents of the A register will now be found to be the original contents added to itself the number of times specified by the multiplier, in other words the product of the two numbers.

DIVISION

To perform division, with the dividend in the A register and the divisor in the E register, first the divisor is transferred to the Z register and then the subtractions are started, each one being counted on the counter. When the contents of the A register go negative (determined by the fact that the borrow store is set at the end of a subtraction the clock is stopped and the quotient will be found to be the content of the counter minus 1.

The minus 1 nuisance is neatly disposed of by presetting the counter to 999999 instead of 000000 before counting takes place, the counter therefore automatically counting the number of necessary sub-tractions minus 1. Finally the quotient stored in the counter is transferred to the A register.

PROGRAMMING AND CONTROL

Up to now the circuits which actually control all the arithmetic and "housekeeping" operations have been ignored, but of course these circuits do have a lot of work to do and are quite extensive. In Digi-Cal the programme is of the wired diode type, each programme being divided up into a series of time periods (eight steps for add and subtract, and sixteen for multiply and divide). Each programme is enabled when the appropriate arithmetic function is selected, and is started by the equals key.

During any one time step, any of the available programme functions can be performed, depending on whether or not a diode is wired in, and this gives a great deal of flexibility in the finer details of the programme which can be extended or altered at will.

The serial additions or subtractions are carried out at very high speed, the clock frequency being in the region of 1MHz, but the programme steps are performed more slowly, making the programme circuits less critical of wire lengths and board layout. The clock pulses themselves are produced by a board housed in the arithmetic section, this board itself being controlled by the programme.

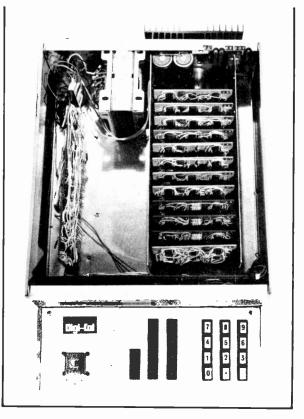
CONSTRUCTION

All the logic used in Digi-Cal is housed on removable boards, the critical arithmetic section on Shirehall DL109 cards, and the display, keyboard and programme on Veroboard panels.

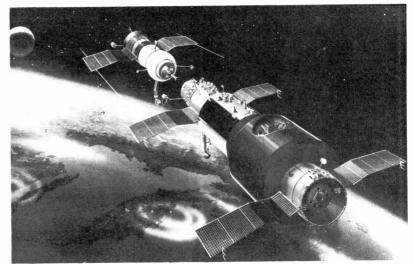
The simple power supplies and regulators are built on a home-etched printed circuit board, the output voltages being five volts for the TTL integrated circuits and 20 volts for the seven segment display.

The keyboard modification to the basic "Contil" case is made with an extra Mod-2 front panel and some $\frac{2}{3}$ in plywood.

Next month: Construction of main chassis and power supplies, and bulk Component List for the complete project







Docking of the Soyuz II and the orbiting scientific station Salvut (picture: Novosti)

SALYUT TWO

The Soviet Union are moving forward with their programme for a successor to the *Soyuz II* and it is expected that a new space laboratory will be launched this year.

It would seem that the design of the new vehicle is more sophisticated and advanced than the American counterpart. This has emerged from the special talks that have been taking place regarding the standardising of docking procedures. Indeed, there has been agreement on a plan for an *Apollo* vehicle to carry out a docking experiment with a *Salyut* vehicle by 1975.

The docking device common to both vehicles will be some 3 metres long, 1.5 metres in diameter and weigh about 2,000 kilogrammes.

The procedure to be adopted is that the Salyut laboratory will first be launched and stabilised in orbit. A day later the crew will set off in a Soyuz vehicle and make their rendezvous. After a lapse of two or three days an Apollo vehicle will attempt a docking with the special device. There is therefore the prospect that three Russian and three American astronauts will orbit the earth together for several days.

The details for the programme are not yet finalised but it could be that both teams will leave the station and return to their respective countries.

No information is at the moment available about the vehicle in preparation but presumably it is much the same as the *Soyus-Salyut* combination of 1971. It consisted of three parts, the service module, the work area and module, and the transfer tunnel. It is to the transfer tunnel that the vehicles come to dock. The scientific experiments are also carried out there.

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The equipment contained in the tunnel includes temperature control, an environmental control system, a sleeping bag with temperature control, general control panels and other scientific equipment. Mounted on the outside of the vehicle are antennas, ion counters, a beacon light, television camera, spherical tanks with air, the solar panels and a telescope for astrophysical studies.

The working compartment houses the more important equipment. This section is of two diameters, the end nearest the access tunnel being about two metres in diameter which opens into a four metre diameter section. The reserves of food and the life support systems are housed here.

Also contained in the larger area are the cooking facilities, power systems, radio and television telemetry, attitude controls, control panels and work areas. There are also the scientific experiments for biology and medical studies together with photographic apparatus.

These ships are crammed with equipment and in spite of there being some 20 portholes they are mostly occupied by some sensing device or other.

The combination of *Soyuz-Salyut* weighs approximately 20 tons. It is about 22 metres long with a maximum diameter of just over 4 metres.

NEW WORLD METEOROLOGICAL PROGRAMME

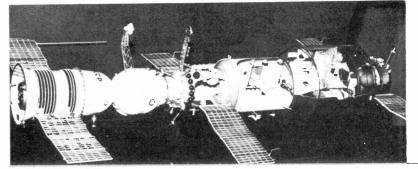
The proposed large scale investigation of the vast energy complexes in certain areas of the world will make use of the techniques of the last decade. The area to be covered is the tropical Atlantic ocean and the adjacent regions of Latin America and Africa. This will enclose a rectangle of more than 260 million square kilometres. This is the tropical region where hurricanes breed.

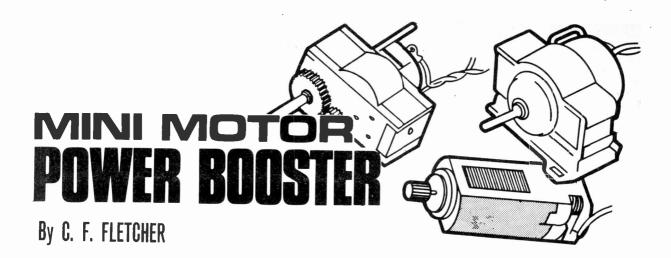
The resources of satellites, ships, balloons, aircraft will be used with the earth stations of the network to feed computers with continuous data. It is hoped that some better understanding of the area will be acquired.

Recent satellite pictures have shown formations which need study for they seem to indicate the manner in which these great turbulences are formed. The sun sucking vast quantities of water into the atmosphere helps to provide the driving force which controls the world wide weather.

A smaller area of something of the order of one million square kilometres near Dakar will also be subject to an intense observation.

Model showing the Soyuz II linked up with Salyut (picture: Novosti)





Over the past few years, toy and working-model manufacturers have made increasing use of the miniature permanent magnet motors originating in the Far East. These "minimotors" are also generally available for model makers and come in two or three frame sizes.

Almost all of them are designed to work from a 1.5 or 3 volt batery and draw around 0.5 ampere when stalled. Looking at these ingeniously made motors one may think of many things to do with them, but not until one has tried something in particular do their shortcomings come to light.

SHORTCOMINGS

The author came face to face with some of these shortcomings when attempting to construct a rotatable indoor television aerial. The motive power to rotate the Yagi array was to be derived from an old electric toy crane drive which used a minimotor with a reduction gear train.

Applying 1.5 volts to this motor caused it to run at a high speed when lightly loaded but it stalled when faced with turning the aerial via a belt drive. Applying 3 volts produced a very good windmill, but very little hope of stopping the aerial in any desired direction.

Something had to be done to reduce the speed of the motor on one hand whilst retaining the high torque provided by the 3 volt supply. Before going on to describe the booster which was designed to do the job, it is worthwhile to consider just why the little motor behaved as it did.

D.C. PERMANENT MAGNET MOTORS

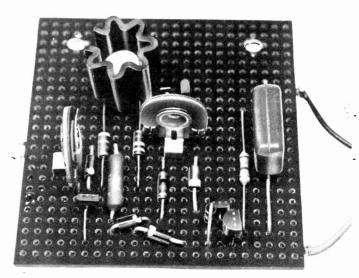
Electromagnetic motors function because of the interaction of two magnetic fields. In the case of a permanent magnet motor, one field is provided by the permanent magnet and the other by the current flowing in the armature windings. The geometry of the motor is such that the force developed between the two magnetic fields drives the armature around and produces the desired rotating action. All very well, but how fast will the motor turn and how much torque (twisting force) will it produce?

To understand this, one needs to consider what happens inside the motor when a battery is connected to the armature. A circuit is made from battery positive, through one of the brushes to the armature windings, then returning to battery negative via the other brush.

The circuit has resistance and so initially a current flows given by

$I = \frac{\text{Battery volts}}{\text{Total circuit resistance}}$

Remember the armature has not yet moved, and so the starting torque is determined by the size of the standstill current which in turn depends upon the circuit resistance and the volts applied.



A photograph of the upper side of the Veroboard panel showing the heatsink mounted on TR4

BACK E.M.F.

Once the motor begins to turn, however, the picture changes, for the movement of the armature windings through the permanent magnetic field causes a voltage to be induced in the armature in opposition to the battery voltage.

The effective voltage is now the difference between these two and the motor reaches a steady speed when the voltage difference drives a current through the armature resistance just big enough to produce the load torque required. No-load requires a small torque, a small current, and a small voltage difference, therefore a high speed is to be expected.

Applying a load slows the armature, increases the voltage difference and causes a bigger current to flow to meet the load. The speed/load characteristic of such a machine is a steadily drooping curve as shown by Fig. 1. Since the minimotors under consideration have a large resistance for their voltage, a sharply drooping load curve results.

Pushing up the applied volts certainly increases the maximum torque but it also makes the no-load speed very high. If the load varies, large variations in the speed must be expected. Thus we have arrived at the inherent disadvantage of these motors: poor speed control and limited starting torque.

ONE SOLUTION

As the armature generated voltage is directly related to armature speed (being zero at standstill), if one could separate this voltage from the armature resistance drop and measure it, one would have a direct means of measuring the speed of the motor.

Knowing the motor speed, a circuit can be designed to constantly monitor it and vary the armature current to keep it steady. Such a "closed loop" system can provide good speed control together with good torque. In fact, by using a supply voltage well in excess of the motor's rated armature voltage, a maximum torque several times normal maximum can be produced.

The circuit to be described uses the well known Wheatstone Bridge measuring system but it has been specially adapted to work from only 4.5 volts and control low voltage motors.

WHEATSTONE BRIDGE

As mentioned earlier, the booster is based upon a Wheatstone Bridge balancing circuit, see Fig. 2. The motor is connected between terminals Y and Z and is represented by R_a , the armature resistance, and E_a , the armature e.m.f.

The bridge function is to separate E_a from the armature resistance drop $I_a \times R_a$. Remember, what appears between terminals Y and Z is $E_a + (I_a \times R_a)$. Applying the Wheatstone principle, if R_3/R_2 is made equal to R_1/R_a then there will be no voltage between X and Y due to the current *I*.

However, when E_a appears due to the armature rotation, this is not balanced out by the bridge and a proportion of it can be measured between X and Y. Since this voltage is directly proportional to E_a , it is also directly proportional to the motor speed.

THE BOOSTER CIRCUIT

Fig. 3 shows the complete circuit diagram of the Minimotor Power Booster.

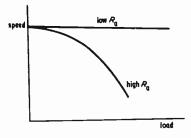


Fig. 1. The speed/load characteristic of a permanent magnet motor

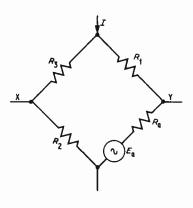


Fig. 2. The Wheatstone Bridge method of measuring the armature e.m.f.

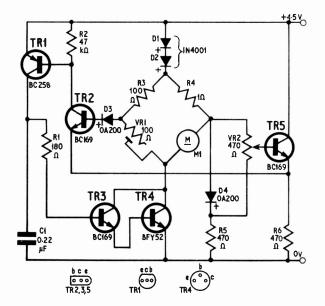


Fig. 3. The complete circuit of the minimotor power booster

COMPONENTS . . .

Resistors	
R1 180Ω	
R2 47kΩ	
R3 100Ω	
R4 1Ω1W	
R5 470Ω	
R6 470Ω	
All \pm 10%, $rac{1}{2}$ W carbon unless otherwise stated	
Potentiometers	
VR1 100 Ω skeleton preset	
VR2 470 Ω skeleton preset or potentiometer (see	
text)	
lexty	
Capacitor	
Ċ1 0·22μF	
Transistors	
TR1 BC258 or 2N4289 or OC202	
TR2, TR3, TR5 BC169 or 2N4291 (3 off)	
TR4 BFY52 or 2N3054	
Diodes	
D1, D2 1N4001 or any 25 p.i.v. 1A diode (2 off)	
D3, D4 OA200 or any small silicon signal diode	
(2 off)	
(2 011)	
Motor	
M1 Any small 1.5 or 3 volt permanent magnet	
motor (see text)	
. ,	
Miscellaneous	
TO5 heatsink for TR4	
Veroboard 0·1in matrix, 2·4in $ imes$ 2·4in	

Current flowing through diode D4 produces approximately 0.6 volts across it which is largely independent of the current. This voltage is used as a reference voltage and part of it is picked off VR2 and applied to TR5 base.

TR5 and TR2 form a voltage comparator which detects any difference between E_a and the reference voltage picked off VR2. If E_a is lower than the reference, current flows in TR2 and is amplified by transistors TR3, 4 and 1. These latter three transistors provide a large current gain and full load motor current can thereby be produced in TR4 with only a small offset between E_a and the reference.

By this action the motor e.m.f. E_a and hence the speed is tied closely to the wanted speed set on VR2.

SPEED CONTROL

The motor used by the author was a 1.5 volt type. If a 3 volt motor is to be controlled, two diodes in series should replace D4 if full speed is required. Again, in the author's model, speed was to be preset. If the constructor requires a variable speed drive, VR2 is simply changed to a normal potentiometer and mounted conveniently.

The supply voltage is not important provided that it is 4.5 volts or more. Be cautious however not to use too high a voltage, because TR4 will have to dissipate more heat as the supply voltage is increased. Mounting TR4 on a substantial heat sink is a worthwhile precaution for a general purpose booster.

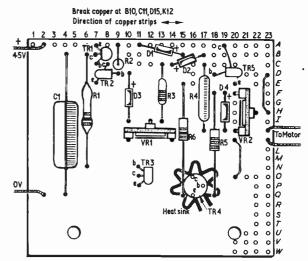


Fig. 4. The layout of the components on the Veroboard panel

For applications where motor current exceeds 1A, use transistor type 2N3054 for TR4.

Diodes D1 and D2 ensure that the comparator has sufficient voltage under all conditions and they pass the full motor current. D3 guarantees the motor will run to standstill when the reference voltage is zero. No special precautions are necessary in construction although short wiring is always to be recommended in high gain amplifiers.

Fig. 4 shows the layout of the components on the Veroboard panel. The preset potentiometer VR2 may be replaced by a potentiometer if variable speed control is required, as mentioned earlier.

CAUTION

Finally a little word of caution. Whereas no damage will be done to the motor, which is made to produce more than its normal power for short periods, sustained overloads will cause overheating and maybe damage. As a general guide, if the motor does not become hot on continuous load or spark excessively at the brushes, it is unlikely to be damaged.

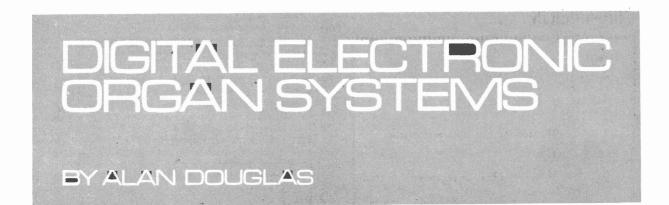
P. F. GEMINI

REPRINTS AVAILABLE

Readers who missed the articles on the "P.E. Gemini" Dual Purpose Stereo Amplifier, published in November 1970 to March 1971, can obtain them in reprint booklet form.

The price of this 32-page booklet is 55p, including postage. Orders for copies, with P.O. or cheque made payable to IPC Magazines Ltd., should be addressed as follows:

The Receiving Cashier (P.E. Gemini) IPC Magazines Ltd., Tower House, Southampton Street, London, W.C.2.



THE electronic organ has passed through more design stages than any other instrument. Many of these were arbitrary since the results are subjective and the requirements of tonal synthesis were not properly understood. With the exception of electro-mechanical generators, all methods used discrete components until the advent of integrated circuits. Printed circuit boards further extend the opportunities for compactness and the last hurdle has now been overcome—the generation of all frequencies from one master oscillator.

This technique enables an organ to be tuned by one control and, since the methods used to derive the required frequency scaling consist of gates, bistables and multivibrators, it becomes possible to vary the tuning within the octave to suit different temperaments—mean tone, just intonation, Werckmeister's, equal temperament, etc. The tuning accuracy is much greater and by employing sufficient binary digits, both this and tonal synthesis can be extended beyond the limits attainable with conventional techniques.

At the outset it must be understood that these advantages are only possible because of microcircuit techniques.

In this article we shall look at digital methods of complete frequency generation from a single oscillator, tonal synthesis and keying control.

SINGLE OSCILLATOR SCALING SYSTEM

The problem of full frequency generation from a single tone source is to find a way to divide the master oscillator frequency by the intervals of a tonal scaling system. The most usual case is that of equal temperament, in which the frequency ratios of any two adjacent notes are in the relationship $1 : {}^{12}\sqrt{2}$, or 1.05946. It is at once obvious that this could not be obtained by a bistable system, it is an irrational number and must be only an approximation. But how good must the approximation be?

One way of solving it is to divide the master oscillator frequency by 196 and then multiply by 185, which is a close approximation to the required value. Any standard organ would be based on C for the top note. Therefore all lower C's can be derived directly from this by bistable division, leaving 11 circuits for the remaining notes of the top octave. Fig. 1 shows the master oscillator entering a divide by 196 multivibrator. As is well known, RC multivibrators readily lock onto an injected frequency even if their free-running frequency is slightly different.

 Master
 Comparator
 Multivibrator

 0scillator
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The output from this circuit now enters a frequency comparator circuit and compares with a divide by 185 multivibrator; the resultant frequency is now B in the equal tempered scale, from which all lower B notes are derived by bistable division in the normal way. Should the division ratios tend to drift, the comparator device holds them to the accuracy required. So to cover the range of notes, there must be a chain of multivibrators and bistables.

TONAL SYNTHESIS

In this, the N. E. Rockwell system, the highest required frequency is set by the demanded accuracy of tonal synthesis. This will be appreciated from Fig. 2 which shows the sampling of an imaginary waveform. Here, the process is additive and the closer the sampling points, the greater the realism.

It has been established that 48 such points will give the maximum realism, this corresponding to 48 harmonics in a normal complex wave system, which is hardly ever attained—certainly not over the whole compass. If 48 points are to be used, then the top frequency must be that of the highest note times 48, for example, top C 4ft = 4,186Hz, therefore 48 \times 4,186 = 200,928Hz.

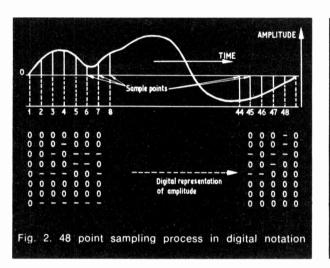
Although it is at once obvious that we can thus provide top C 2ft (8,372Hz and above), there would not be the same number of sampling points; and this is rational, since most of these would be far

above audibility, although it is known that beats between audible and inaudible frequencies do occur and do modify high pitched tones to some extent.

DIGITAL CONVERSION

So far we have provided an array of frequency sources, but these can be used to cover full tonal synthesis by digital conversion. Looking again at Fig. 2, we see that sampling points are represented by digital representation of level in the form of 8 bit words to which are assigned binary code numbers. sustained part of the signal (i.e. as long as the key is held down) and a somewhat similar decay so that the sound does not stop abruptly.

The shift registers are operated from the ring counters which assign the chosen frequencies to the diode matrix gates, and are so arranged that each count shifts a digital word one position to the right; this corresponds to an amplitude division by two, hence by using any desired number of register stages, the amplitude is decreased in steps to zero level. Although this decay is in step form, the ear hears them only as a smooth release of the tone, as seen in Fig. 3.

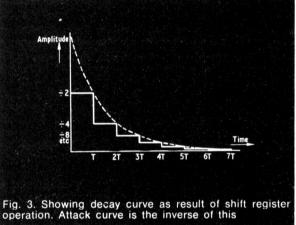


If a C_1 key is depressed, the frequency synthesiser will energise a ring counter which will repeatedly produce an output of $C_1 \times 48$ into a diode storage matrix where all the amplitude increments are stored in binary form. If all are connected for this note, 6 bit signals corresponding to the stored digital words will appear at the output at a rate of $48 \times C_1$ words per second, thus the sampling rate is the same but the actual frequency will vary according to the key selected. The number of diodes connected will be controlled by the stops, which operate gates connecting the composite frequency outputs to the keying system.

KEYING CONTROL

In conventional keying, attack, hold and decay are usually provided by rather crude RC time constant circuits, unless the organ if of the free-phase type with independent oscillators where these attributes can be individually determined for each note. However, it is generally conceded that if the rate of attack is equal to the duration of about 7 periods of the frequency in question, the audible effect is like that of an 8ft organ pipe.

In the digital system a one-shot multivibrator having a period of 7 cycles is used in conjunction with a shift register to provide a series of logarithmic steps simulating the rise in sound level, the



CONVERSION

Subsequent to the control of the required number of notes played (e.g. a chord), a summing amplifier adds all the signals from the decay circuits and since they are still in digital form, they must be turned into an audio signal by a digital to analogue converter. The input to this consists solely of step functions, which will have a high-frequency component; this may be left in to simulate wind noise which of course is a component of all organ flue pipes, and without which true synthesis is not possible.

Many modifications to the basic principles outlined are possible as, for example, exhibited in the N. E. Rockwell organ system which is used in the Allen organ.

PHILIPS SYSTEM

The next system to be looked at is the one devised by N.V. Philips, of Eindhoven, Holland, This is a digital organ generator which produces frequencies by a subtractive method—rejecting a certain number of pulses from a series of pulses.

The number to be rejected will be $2^{-1/12}$ times the number of input pulses. Twelve such circuits will then form a tempered octave, but because of the irrational divisor the output pulses will be spaced



AC127 AC128	17p AF125 17p AF126	20p BCI47 20p BCI48	17p BCY71 12p BCY72	30p BF316 15p BFW10	75p MAT121 55p MPF102	17p UT46 43p V405A	27p 2N1304 25p 2N1305	20p 2N2907 20p 2N2907A		9 40p
ACI4IK	17p AF127	20p BC 49	17p BCZ11	20p BFX29	27p MPF105	43p V410A	45p 2N1306	22p 2N2923	13p 2N3820	3 25p
ACI42K	17p AF139	33p BC 50	17p BD121	85p BFX84	20p OC19	30p 2G301	19p 2N1307	22p 2N2924	13p 2N3903	
ACI5I	15p AF178	50p BC 5	20p BD123	85p BFX85	27p OC20	50p 2G302	19p 2N1308	27p 2N2925	13p 2N3904	
AC154	15p AF179	50p BC152	17p 8D124	75p BFX86	22p OC22	30p 2G303	19p 2N1309	27p 2N2926	2N3905	5 25p
AC155	17p AF180	50p BC153	27p 8D131	80p BFX87	25p OC23	33p 2G304	20p 2N1613	17p (G)	12p 2N3906	5 27p
AC156	17p AF191	50p BC154	30p BD132	80p BFX88	22p OC24	45p 2G306	35p 2N1711	20p 2N2926()	r) IIp 2N4058	€ 10p
AC157	17p AF186	45p BC157	20p BDY20	£1 BFY50	20p OC25	25p 2G308	35p 2N1889	35p 2N2926	2N4059	
AC165	17p AF239	37p BC158	17p BF115	22p BFY51	20p OC26	25p 2G309	35p 2N1890	45p (O)	10p 2N4060	
AC166	17p AFZ11	37p BC159	20p BFI17	45p BFY52	20p OC28	40p 2G339	17p 2N1893	37p 2N3010	80p 2N4061	2p
AC167	20p AFZ12	45p BC167	13p BF118	60p BFY53	17p OC29	40p 2G339A	15p 2N2160	60p 2N3011	20p 2N4062	≥ 2p
AC168	20p AL102	85p BC168	13p BF119	70p BSX19	15p OC35	33p 2G344	15p 2N2147	75p 2N3053	20p 2N5172	
AC169	14p AL103	85p BC169	13p BF152	35p BSX20	15p OC36	40p 2G345	15p 2N2148	60p 2N3054	50p 2N5459	
AC176	23p ASY26	25p BC170	12p BF153	35p BSY25	15p OC41	20p 2G371	13p 2N2192	30p 2N3055	63p 2S034	
AC177 AC187 AC188	20p ASY27 30p ASY28 30p ASY29	30p BC171 25p BC172	13p BF154 13p BF157	35p BSY26 45p BSY27	15p OC42 15p OC44	22p 2G371B 15p 2G374	10p 2N2193 17p 2N2194 27p 2N2217	30p 2N3391 27p 2N3391A 20p 2N3392		
ACY17 ACY18	25p ASY50 20p ASY51	25p BC173 25p BC174 25p BC175	13p BF158 13p BF159 22p BF160	25p BSY28 30p BSY29 30p BSY38	15p OC45 15p OC70 15p OC71	12p 2G377 15p 2G378 9p 2G382	15p 2N2218 15p 2N2218	25p 2N3393 27p 2N3394	15p 25303 15p 25304	45p 60p ≰l·l0
ACY19 ACY20 ACY21	22p ASY52 20p ASY54 20p ASY55	25p BC177 25p BC178 25p BC179	17p BF162 17p BF163 17p BF164	30p BSY39 35p BSY40 35p BSY41	15p OC72 30p OC74 35p OC75	12p 2G401 12p 2G414 15p 2G417	30p 2N2220 30p 2N2221 25p 2N2222	22p 2N3395 22p 2N3402 27p 2N3403	20p 25305 22p 25306 22p 25307	£1-10 £1-10
ACY22	19p ASY56	25p BC180	20p BF165	35p BSY95	12p OC76	15p 2N388	30p 2N2368	17p 2N3404	32p 25321	60p
ACY27	18p ASY57	25p BC181	22p BF167	22p BSY95A	12p OC77	25p 2N388A	50p 2N2369	15p 2N3405	45p 25322	50p
ACY28	19p ASY58	25p BC182	10p BF173	22p BU105	23-90 OC81	15p 2N404	22p 2N2369A	15p 2N3414	20p 25322A	45p
ACY29	30p ASY58	25p BC182L	10p BF176	35p CIIIE	60p OC81D	15p 2N404A	30p 2N2411	50p 2N3415	20p 25323	60p
ACY30	25p ASZ21	40p BC183	10p BF177	35p C400	30p OC82	15p 2N524	55p 2N2412	50p 2N3417	37p 25324	£l·20
ACY31 ACY34 ACY35	25p BC107 18p BC108 18p BC109	10p BC183L 10p BC184 11p BC184L	10p BF178 13p BF179 13p BF180	45p C407 50p C424 30p C425	25p OC82D 17p OC83 40p OC84	15p 2N527 20p 2N696	60p 2N2646 12p 2N2711 15p 2N2712	55p 2N3525 22p 2N3702 22p 2N3703	74p 25325 12p 25326 12p 25327	£1-20 £1-20 £1-20
ACY36 ACY40	30p BC113 15p BC114	25p BC186 30p BC187	27p BF181 27p BF182	30p C426 30p C428	30p OC139 20p OC140	20p 2N697 15p 2N698 17p 2N699	24p 55p DIC	DES & R		
ACY41 ACY44 AD140	18p BC115 35p BC116 40p BC117	30p BC207 35p BC209 35p BC209	p BF183 p BF184 p BF185	30p C441 25p C442 30p C444	27p OC170 35p OC171 37p OC200	15p 2N706 15p 2N706A 25p 2N708	7p 8p AA119 12p AA120	8p BYZII 8p BYZI2	32p OA81 30p OA85	7р 7р
AD142	40p BC118	25p BC212L	11p BF188	30p C450	17p OC201	27p IN709	45p BA116	22p BYZI3	25p OA90	6p
AD149	43p BC119	45p BC213L	11p BF194	23p C720	12p OC202	27p 2N711	40p BA126	22p BYZI6	35p OA91	7p
AD161	35p BC125	35p BC213L	11p BF195	24p C722	25p OC203	25p 2N717	42p BY100	15p BYZI7	35p OA95	7p
AD162	35p BC126	35p BC214L	12p BF196	30p C740	25p OC204	25p 2N718	24p BY101	12p BYZ18	30p OA200	6p
AD161/	BC132	25p BC225	25p BF197	35p C742	17p OC205	35p 2N718A	50p BY105	15p BYZ19	25p OA202	7p
162(MP)	63p BCI34	30p BC226	35p BF200	45p C744	17p OC309	35p 2N726	27p BY114	12p OA5	17p SO10	4p
ADT140	50p BCI35	30p BC317	12p BF222	80p C760	17p P346A	17p 2N727	27p BY126	15p OA10	22p SO19	4p
ADZ11	62 BCI36	30p BC318	12p BF257	35p C762	17p P397	45p 2N743	17p BY127	17p OA47	7p IN914	6p
ADZ12 @	2-10 BC137	35p BC319	12p BF270	25p C764	60p OCP71	43p 2N744	17p BY130	15p OA70	7p IN916	6p
AF114	17p BC139	45p BCY30	20p BF271	17p EC401	15p ORP12	43p 2N914	17p BYZ10	35p OA79	8p IN4148	6p

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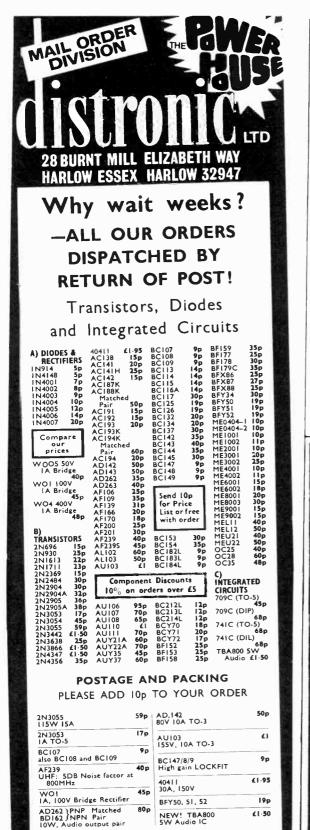
TYPE ST N5. Silicon Planar Transistors npn TO-5 Metal Can. Types similar to: BFY50-SI-52 and 2N2192-92. Price: 500 £9-50; 1,000 £16

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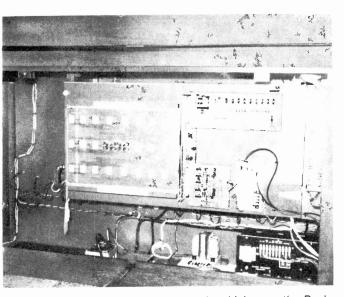
Price: 500 £7-50; 1,000 £13 TYPE STNK. Silicon Planar Plastic Transistor npn with TO-18 pin circular lead configuration, I.C. 200mA, 300mW and similar to BC1078-9, BC170, BC173, BC182-184, BC237-8-9 and BC337-8. Price: 500 69:50; 1,000 £16

When ordering, please state type required, i.e., STNK or STN18, etc.









Photograph of the Allen console which uses the Rockwell organ system. This measures about 41ft in length

Table 1.

RATIOS OF FREQUENCIES WITHIN THE OCTAVE

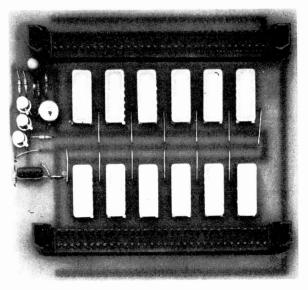
Note	Decimal	Binary
C'	2.0000	10.0000000000
В	1.8878	1.1110001101
Α	1.7818	1.1100100001
A G	1.6816	1.1010111010
G	1.5874	1.1001011010
G	1.4983	1.0111111110
F	1.4142	1.0110101000
F	1.3348	1.0101010111
E	1.2599	1.0100001010
D	1.1892	1.0011000010
D	1.1225	1.0001111101
С	1.0595	1.0000111101
С	1.0000	1.0000000000

irregularly, making the resultant tones sound harsh. To remedy this, the frequency is made very much higher than the desired values, and the output frequencies are divided by the same factor before using them for tone forming. Since the required frequencies will be related to the equal tempered scale, they must be derived digitally.

We have just seen one approach, and other investigators have tried to obtain the exact intervals by successive division by large integers between 200 and 300. But in this case, the required number of pulses is obtained by adding trains of pulses at different repetition rates.

BINARY DIVISION

In Table 1 we see the binary notation corresponding to one octave. The master frequency is 10 and the half frequency is 1; all notes within the octave must be given by a binary number between 1 and



Philips' digital organ generator

10. Pulse trains obtained by halving the master oscillator frequency (suitably reduced by division) correspond successively with the binary numbers 0.1, 0.01, 0.001, etc. and provide the material for making up the binary numbers between 1 and 10 with the required accuracy (no deviation greater than 0.05 per cent = $1/2.000 = 1/2^{11}$, hence 11 binary numbers). If the frequency 1 is note C, then A is formed by adding the following pulse trains:

1
0.1
0.001
0.00001
0.000001
0.0000001
0.000000001
1.101011101

Other notes can be formed in a similar way, but it has been found easier to form the note by subtraction from the binary 10 of the primary frequency. The A just mentioned is equal to 1.1010111010 which is equal to 10-0.0101000110.

This month's cover picture was photographed in the "Bird's Nest" pub in the Kings Road, Chelsea, London.

An Index for Volume Seven (January 1971 to December 1971) is now available price 10p inclusive of postage.

BINDERS

Binders for P.E., with a special pocket for storing booklets and data sheets, are available price $\pounds 1$, including postage and packing. State Volume Number required.

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A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.

This is YOUR page and any idea published will be awarded payment according to its merits.



As the number of constructional projects employing i.e.s is increasing, this simple logic probe may prove invaluable to fault finding in such circuits. It consists of a 932 (or 944) DTL i.e. and a 6V 0 06A lamp, see Fig. 1.

As only half of the i.c. is used it is possible to salvage a partially functional device which would otherwise have little value in circuit construction.

The probe, Fig. 2, can be constructed on a ballpoint pen barrel. In operation the lamp will light when a "1" logic level is detected, and will be extinguished when a "0" level is detected. If a continuously pulsed input is applied the lamp will glow dimly.

Any of the inputs may be used as shown, except for inputs 4 and 11 which require a silicon diode in series with the probe lead.

> A. Meldrum, Fife, Scotland.

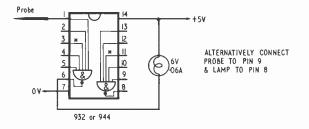


Fig. 1. Circuit details of the logic probe

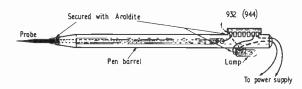
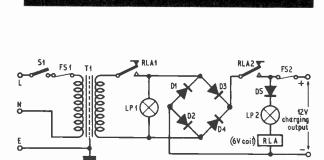


Fig. 2. Constructional details of the probe



BATTERY CHARGER PROTECTION

Fig. 1. Battery protection circuit diagram

A BATTERY charger, consisting of nothing more than a transformer and a bridge rectifier, was found to be non-operational and enquiries showed that the output connections had been accidentally reversed.

As explanations about two conducting diodes in series and no current limiting resistance just do not register with everyone, I decided on the modification shown in Fig. 1. I don't say it's completely foolproof, but it can anticipate most of the common evils perpetrated by the non-technical.

The normal charger transformer and bridge circuit is broken by two relay contacts, one in the a.c. supply to the bridge RLA1 and one in the d.c. output line from the bridge RLA2. The relay coil, diode D5 and lamp LP2 (acting as a ballast resistor) are wired in series and then connected across the output of the charger, see Fig. 1.

If the charger is plugged into the mains with the output leads shorted together nothing happens because the relay contacts have not closed and there is no a.c. supply to the bridge; neither pilot light will light. If next, the charger is connected to the car battery the wrong way round nothing will happen because the diode in series with the relay coil prevents it from energising; neither lamp lighting.

Only by connecting the charger to the battery the right way round will the relay pull-in and the pilot lamp LP2 light to indicate that the battery is connected correctly. Having connected the battery correctly, the relay is energised and allows a.c. to flow to the bridge via RLA1; LP1 will now be illuminated. When both lamps LP2 and LP1 are illuminated it indicates that the charger is functioning correctly.

The relay used can be any type which will pull-in at six volts, and has two "make" contacts capable of handling the output charging current to the battery. Ideally, the relay coil current should be 300mA, or as near to the current required by the pilot bulb LP2 in series with it as possible. Likewise, the diode D5 in series with the lamp is not critical; choose one that has about double the current capacity required, and will stand approximately 50V peak inverse voltage.

G. Read, Southampton, Hants.



Practical Electronics July 1972

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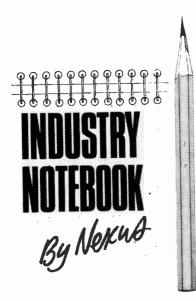
OPTIMISM IN THE AIR

A TTENDANCE at the giant IEA exhibition at Olympia was expected to be somewhat down. This will surprise few people, although optimism for the trading future was up, even in the face of all the trouble on the railways and the docks.

As I reported in this column last month, the Paris Components Show, first trade barometer of the year, showed an upward trend in confidence.

After Paris I visited Denmark to get a line on Danish sentiment. Industry leaders I met all admitted that 1971 was a bad year but the universal opinion was that this year will see the turnround.

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The Danes are doing well in electronics through being very very good in comparatively narrow sectors of the industry. A country of less than five million people can't be good in every sector and to get to a viable operational size in world markets the Danes have had to go for big exports. The two top instrument companies, Brüel and Kjær, and Radiometer, both export over 95 per cent of their production. Don't underestimate the Danes. They are good engineers in their specialities and know how to sell.

The U.K. is a big investor in Denmark with Rank owning Rank-Arena, the big producers of Danish radio and television, and EMI owning the largest chain of radio retailing outlets. And UK-Danish electronic imports and exports have greatly expanded during the past five years.

Overall, I find my many inputs of information, not only from the U.K. but also from Europe, almost universally optimistic.

BRAVE EXPERIMENT

Seminex, the first seminar and exhibition organised specially for the semiconductor industry was a brave experiment which deserved a better fate than to be hit by the rail dispute which strongly affected attendance. Even though the event must be counted as a qualified success, its organiser, Evan Steadman, is not letting the grass grow under his feet.

As soon as Seminex was over he mailed a questionnaire to all and sundry. He honestly admitted it could have been better. Could he kindly have comments and suggestions?

Evan Steadman is a go-getter not particularly noted for humility. But he is determined to make Seminex an event on the British exhibition calendar and is now planning to take Seminex on tour to the European mainland and, perhaps, even further afield. We wish him well.

COMMUNICATION 72

Another brave experiment, well patronised by the industry, is Communication 72 held at Brighton over the three days June 13-17.

It has the full support of the Electronic Engineering Association and the Department of Trade and Industry in bringing in three trade missions from overseas who are showing the British communication industry in action as well as sitting in at the Brighton seminar and looking at the exhibition. Each mission consists of about 15 senior officials from abroad, all users of radio and data systems. The Conference organisers are IPC Electrical and Electronic Press Ltd.

Communication 72 is yet another example of the trend towards smaller, more specialist exhibitions. But the traditional giant shows are taking a long time to die.

NEW ON THE SCENE

Two businessmen are about to re-emerge with a new commercial image. First, ex-Avo sales director John Minister should make a success of Risso Electronics Ltd, which is just moving off the ground as a new competitor in the busy instrument market.

Risso (what about "The Risso Sign means Happy Metering" for a slogan?) has just won its first major contract from a public corporation for a multi-range meter. The production line is already in action. Risso has also designed a range of panel meters.

Secondly, Keith Harris, ex-marketing director of Keyswitch Relays, comes up as a new force in the relay business. This, I understand, will be a marketing operation of a nature as yet undisclosed but while being reticent about his new business venture he is nevertheless full of confidence for the future.

THE TAKEOVER TRAIL

One of the fascinations of the industry is its dynamism, not only in technology but also in its business structure. After every big merger there is a spawning of little companies. Lots of people don't like working for the big corporations and, more often than not, after a big merger there is an inevitable "shake out" of what is loosely termed middle management, men who are willing and in some cases anxious to strike out on their own.

But it works the other way, too, with plenty of smallish companies seeking the shelter (with a cash handout) of a larger grouping. And, of course, a move in this direction can make a lot of sense, especially if the product lines are complementary.

The group to watch for real finesse in acquiring complementary companies is Racal whose chairman and managing director, Ernie Harrison, has a keen nose for this sort of thing.

Mum's the word during negotiations but yet another company, whose name may be announced before this column is printed, is about to join the Racal Group. It will be the third Racal acquisition in less than twelve months and my guess is that it won't be the last this year.

INSTRUMENT'S TRAVAIL

The UK has no instrument company in the world class and by this I mean companies like Hewlett-Packard or Tektronix. Many leaders in the industry regret this, especially as instrument sales have been stagnant now for three or four years.

The solution would be in regrouping into larger units but though there have been discussions on the possibility, no common ground of agreement has been found.

The differences have centred on who is to be the nucleus company. There are plenty of possibles but while all are prepared to be the nucleus, none are willing to be the satellites. And can you really blame them?

Meantime, one British instrument company at least is marketing, under its own brand, instruments made in Japan and Sweden to name but two countries who are getting a share of the British market which rightly should go to our own manufacturing units.

It's time these differences were resolved.

More on LOGICAL RADIO CONTROL

THE short series of articles Logical Radio Control published earlier this year (reprints and back numbers are not available) has provoked much correspondence, some of which has been published in *Readout*. Some of the more general questions with the author's replies are given here together with further details on the Decoder module.

SYNC PULSE GENERATOR

Does one need to use the F9601 retriggerable monostable? Can an alternative circuit be used and could a substitute be found from the Texas SN74 range?

Many alternative forms of sync generator are possible, some being quite economic. Having regard to the convenience of fitting only *one* i.c., its reliability and the question of noise immunity, the retriggerable monostable technique is recommended.

Excessive noise, for example, interference from a servo-motor, must not be allowed to give spurious sync pulses leading to complete breakdown of control.

If the specified i.e. cannot be obtained conveniently the SGS type T118D1 can be used. This is a direct replacement type and is available from Quarndron Electronics (Semiconductors) Ltd., Slack Lane, Derby DE3 3ED, price £2.70.

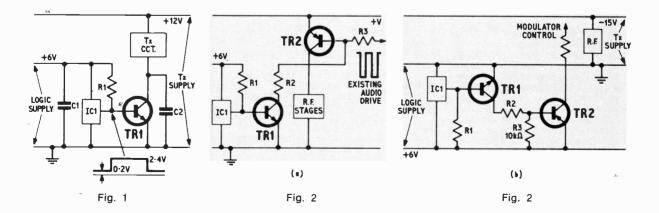
The Texas SN74122 may be used by comparing the manufacturers' information and changing connections accordingly. This unit is priced £1.44.

INTERFACE UNITS

Can the Coder 2B output be used to switch the modulator of a transmitter and if not can you suggest a circuit?

In general the output voltage of an integrated circuit unit is either a minimum of 2.4V in the high or logical "1" state or is a maximum of 0.4V (typically 0.2V) in the low or logical "0" state.

The most convenient and definite interface stage is one using a silicon *npn* transistor (e.g. 2N3704); this is held in the "off" condition by a logical "0" output of 0.4V, which is less than the "turn on" voltage $V_{\rm he}$ of 0.6V. In the logical "1" state the transistor is turned on and an additional resistor R1, as shown in Fig. 1 (of value greater than 5001), to prevent overloading the i.c.) is used to supply adequate bias current to bottom or saturate the transistor.



In the case of a transmitter with a positive supply (with respect to earth) the units are connected as shown in Fig. 1.

In the "off" state the collector of TR1 rises to the voltage of the transmitter supply (say + 12V) and when "on" or bottomed is at approximately +0.2V. Therefore, TR1 acts as a supply switch, with a loss of 0.2V and may be used to control either the r.f. output stage or preferably, the modulating transistor or stages, avoiding difficulties of r.f. decoupling and pulse shaping.

The selection of TR1 (Fig. 1) ensures that:

(a) The maximum current rating (800mA for 2N3704) is not exceeded by the load of the controlled stage, including the charging current of the electrolytic capacitors used for decoupling. interfacing, necessitating only that an inversion, using one transistor or a gate, is used in the receiver, so that zero modulation corresponding to a coder pulse, produces a positive going input to the decoder.

The value of resistor R2 is chosen to bottom TR2. If TR2 has a nominal "gain" of say 50 and normally passes 50mA to the load, a base current of approximately 2mA is necessary to ensure bottoming. For a 12V supply

 $R_{2} = 12 - (V_{ce(TR_{1})} + V_{be(TR_{2})}) \times \frac{1,000}{2} \text{ ohms}$ $= 12 - (0.7 + 0.3) \times \frac{1,000}{2} \text{ ohms}$ $\simeq \frac{11,000}{2} \simeq 5,500 \text{ ohms}$

or to the nearest preferred value $-5.1 k\Omega$.

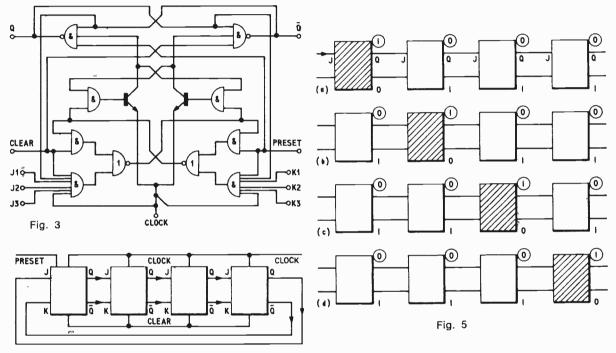


Fig. 4

(b) The maximum collector open circuit voltage $(V_{vo} = 50V \text{ for } 2N3704)$ is not exceeded bearing in mind that an undecoupled inductive load, such as the r.f. "tank" coil, produces collector voltages several times that of the supply. R.F. decoupling capacitors (0.1 to $0.01\mu\text{F}$) should be fitted where shown at C1, C2.

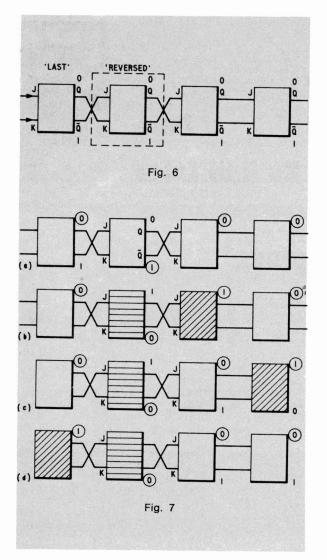
USUAL DESIGN

A usual design for simple transmitters is shown in Fig. 2a where the r.f. output npn transistor is in series with a pnp switching transistor driven from an audio multivibrator circuit. In this case the action of TR1, when cut off, is to allow modulation, but when "on," TR2 is bottomed giving full unmodulated r.f. output. This is a convenient method of In cases where the transmitter uses a negative supply the circuit shown in Fig. 2b should be used. Here TR1 (npn) is turned on by a logic "1" as before, providing a bias current via R2 to saturate TR2 (pnp) which should be selected as before, to perform the required switching action, bearing in mind that the effective supply to TR2 is the transmitter supply voltage plus the logic supply.

DECODER

Do the Decoder outputs go high after the pulses and remain high until the next clear pulse?

For purposes of illustration only Fig. 3 shows the internal functions of an SN7472 JK master-slave flip-flop. The units used in the article are the dual



version of this unit, i.e. two inside one integrated circuit package, and the preset input is not available.

Though apparently very complicated, they are simple to use necessitating only a knowledge of this simplified truth table:

Befo	re Clock	After Clock pulse
inputs		Q output
J	K	
0	1	0
1	. 0	1

The \overline{Q} output is the inverse of the Q output and the sequence of operation is

- 1. Isolate slave from master;
- 2. Enter AND gate information to master;
- 3. Disable AND gate inputs;

4. Transfer information from master to slave.

Also, preset (which sets Q to 1) and the clear (which sets \overline{Q} to 1) inputs operate on negative-going signals.

FLIP-FLOP RING

Consider a "ring" of any number of such flip-flops connected as shown in Fig. 4 with each Q output connected to the next J input and \overline{Q} output to K inputs.

When power is applied the units may take up any random position but a subsequent clear pulse will set Q outputs to 0 and all \overline{Q} outputs to 1. Any subsequent clock pulse will not change the situation since each J input "sees" a "0" from the Q output of the preceeding stage and similarly the K input is a "1."

From the truth table the output cannot change: if however a preset pulse is applied to only one unit making its Q output a "1" as shown in Fig. 5 a subsequent clock pulse will cause the next flip-flop to change state since its "J" input sees a "1." The flip-flop that was preset will also change back to a Q = 0 output since its J input "sees" a "0."

Further clock pulses will cause successive flip-flops to change state and back again so that the "1" (shown shaded) moves round the ring of Q outputs.

The Q = 1 or high output of an individual flipflop can be identified therefore with a particular clock pulse and only when, say, number 3 clock pulse is present, will the corresponding number 3 flip-flop output be high.

As mentioned in the article, the flip-flop output is high for the duration of the clock pulse and for the brief and defined period of inter-clock pulse "gap" of $\frac{1}{2}$ to $\frac{1}{2}$ millisecond. The flip-flop output is integrated to give a d.c. level dependent upon the duration of the pulse; the small additional "gap" period is included in this, appearing as a small constant d.c. level off-set by initial adjustment of the servo. Any small time-by-time variation of the inter-clock pulse gap (say 5% of 0.5ms) is insignificant with respect to the cycle time of about 25ms.

REVERSED OUTPUT

To avoid the necessity of providing a preset pulse, after the clear and prior to the train of clock pulses (particularly since low-cost i.c.s with multiple flipflops such as the SN7473 do not have a preset input) the ring of flip-flops is arranged as shown in Fig. 6 with the input and output connections of one flipflop reversed.

Since the "J" input of the following unit now sees a "1" from the \overline{Q} output, it changes state on the first clock pulse. The "reversed" flip-flop also changes state since its J input sees a 1 from the \overline{Q} output of the "last" flip-flop.

Further clock pulses cause the "1" to move round the ring of Q outputs as before (shown in Fig. 7). Subsequent to the "last" flip-flop changing state, the "reversed" flip-flop will change back to the initial condition.

The "reversed" flip-flop is used therefore to store the necessary "1." The normal flip-flop Q outputs correspond to the clock pulse-channel outputs as before except that the \overline{Q} output of the "reversed" flip-flop can be used as explained in the article, using gates, to provide a channel output.

In the wiring pattern in Fig. 20 (January 72 issue) a double reversal appears in error. To correct this the connections to pins 8 and 9 of IC2 (at the bottom centre) should be reversed.

ALPHA NUMERIC Part DISPLAYS By R.W.Coles



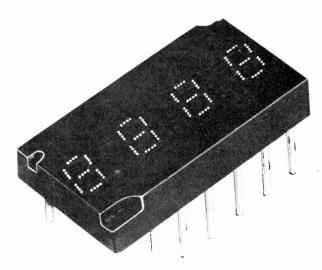
Light Emitting Diode Displays

THE ALPHA-NUMERIC display devices considered so far in this series (gas discharge and filament types), have utilised basic principles which have been known for many years and which have been employed for other purposes before their incorporation in display systems.

This month is concerned with the brand new technology of solid state semiconductor light emitters and the incorporation of this technology into readout devices.

SEMICONDUCTOR LIGHT EMISSION

Silicon and germanium are well known as semiconductors, but they are by no means the only substances which bear this label, and it was while investigating the properties of some of the more exotic members of the family that it was discovered that certain semiconductors have the property of emitting radiation when a current flows through them.



Each segment in this device is made up of four light emitting diodes (Guest International)

The first practical device to result from this research was the gallium arsenide (GaAs) diode, which, when forward biased, emitted infra-red radiation.

This type of device found many uses in speechlink and beam-breaking equipment, and has been featured several times in this magazine.

The radiation from L.E.D.s is monochromatic, that is, it has a single frequency or colour, and, since it is in the infra-red range, the radiation from a GaAs diode is not visible to the naked eye and is therefore unsuitable for display purposes.

Further research produced a diode using gallium arsenide phosphide (GaAsP) which emitted light in the red portion of the visible spectrum, making it ideal for incorporation in readout devices, and it is diodes of this type which will be dealt with in this article.

A very informative account of the exact mechanism involve in L.E.D. structures was contained in the article on *Electroluminescence* in the December 1971 issue, and it is beyond the scope of this article to delve any further into this aspect of the subject.

DECREASING PRICES

The fact that L.E.D.s are made in the same way as other semiconductor devices such as transistors and integrated circuits means that they are set fair for mass production by basically standard methods, and already the explosion of these new indicators into the display market is beginning.

Prices, which were prohibitively high only a couple of years ago, are now dropping rapidly in the same way that i.c. prices did, and within a year or two there is little doubt that L.E.D. indicators will have the lion's share of the display market.

CONSTRUCTION OF L.E.D. DISPLAYS

Alpha-numeric readouts using L.E.D.s employ numbers of square or oblong emitting areas arranged either as a "dot matrix" or a "bar matrix." They are normally laid out on a single slice of semiconductor material, the whole chip being enclosed in a package like an integrated circuit, except that the packaging compound is transparent rather than opaque.

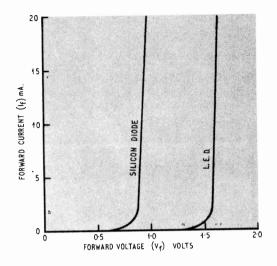


Fig. 5.1. Forward characteristic of a typical L.E.D. compared with a silicon diode

Because L.E.D. readouts are made by manufacturers specialising in other semiconductor devices it is a logical step to include some or all of the decoding and driving electronics for the display inside the same package as the L.E.D.s in the form of an i.c. chip. This is a possibility not open to other display technologies, and it is a very important advantage which, no doubt, will eventually be exploited to the full.

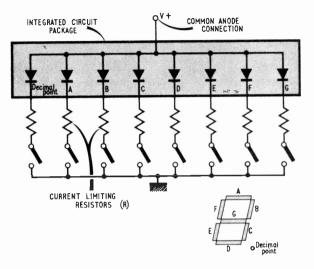
Perhaps the most striking advantage of these new devices is their small size, character height commonly being in the region of only 1 in, making it possible to enclose up to six separate digits in a single dualin-line package, for example.

Other advantages include very long life; freedom from catastrophic failure; compatibility with all types of drive circuitry; and most important, the ability to be run from a wide variety of supply voltages from 2V upwards.

DRAWBACKS

Of course, there are a few drawbacks, the small physical size quoted above as an advantage can be a disadvantage for displays which are to be read from more than a few feet away. Large L.E.D. arrays are not practical in integrated form due to the large area of semiconductor which would be necessary, and these devices are not likely to be competitive when character height exceeds $\frac{1}{2}$ in.

The monochromatic red colour, whilst ideal for calculators and measuring instruments is undesirable in such applications as film annotation where certain emulsions have a blind spot in this region of the spectrum. It is worth mentioning that this colour problem has been overcome by using a different semiconductor, and both green and amber displays have been produced in the laboratory, but as yet the price of these devices is higher than equivalent red light types.



1

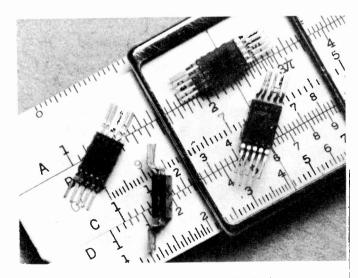
Fig. 5.2. Basic circuit arrangement for a sevensegment L.E.D. indicator. The switches shown would normally be replaced by the output stage of an i.c. decoder/driver

DEVICE CHARACTERISTICS

A single L.E.D. behaves in a similar way to the silicon and germanium components which have become so familiar.

Its forward characteristic is shown in the graph of Fig. 5.1, where it is compared with that of a silicon diode. As can be seen, the only difference is in the voltage at which it begins to conduct, being typically 1.6V for an L.E.D. as against 0.8V for a silicon device.

In a practical display a single L.E.D. is used for each bar or dot of the matrix, though in the larger seven-bar types it is necessary to use two (or sometimes more) L.E.D.s in series for each bar which



These L.E.D. indicators from Mullard (type 185CQY) show an alternative method of packaging to the normal D.I.P. type

gives an apparent doubling of the forward voltage necessary for conduction to typically 3.2V.

To drive an L.E.D. display it is only necessary to insert a resistor in series with each separate element to limit the current to the desired value (usually between 5 and 10mA). The circuit arrangement for a single digit using a seven segment device is shown in Fig. 5.2. In practice the switches shown would be replaced with the output drivers of a seven segment decoder such as the SN7447 discussed last month.

LIMITING RESISTOR

The value of the resistor required can be worked out very easily by working out the voltage across it and dividing it by the required current. The voltage across the resistor is equal to the supply voltage minus both the voltage across the L.E.D. (1.6 or 3.2V as discussed earlier), and any saturation voltage of the switch (around 0.2V for the SN447 i.c.).

For example, if the desired current is 5mA and each bar of the display consists of a single diode, with an available supply voltage of 5V the value of R is found as follows:—

$$R = \frac{V_{CC} - V_{LED} - V_{CE(Sat)}}{I_{LED}}$$

= $\frac{5 \cdot 0 - 1 \cdot 6 - 0 \cdot 2}{5}$ kilohms (where I_{LED} is in mA)
= $\frac{3 \cdot 2}{5}$ kilohms

Thus R = 640 ohms (620 ohms preferred value could be used).

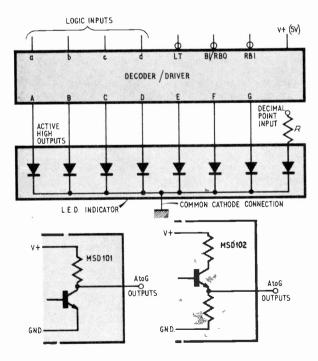


Fig. 5.3. Circuit arrangement for a common cathode type L.E.D. indicator. Also shown are the output stages of two types of driver i.c.s, the Monsanto MSD101 and MSD102

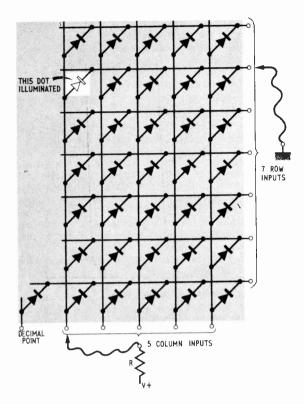


Fig. 5.4. The internal circuit of an L.E.D. indicator using a 5×7 matrix of diodes. The external components show how a particular diode is illuminated though the switching in a real system would be done electronically

COMMON CATHODE

The seven bar indicator just considered had a common connection to all seven diode anodes, the single pin used to carry this common lead being connected externally to the positive supply. It is equally possible to have a common cathode connection which is taken to the negative supply, an arrangement which is suited to a new generation of decoder/drivers which have emitter follower outputs and resistors included in the package, see Fig. 5.3.

DIODE ACTION

The fact that an L.E.D. is not just a light source but also a semiconductor diode can be used to great advantage where time sharing of a display is necessary, since the diode action can be used as part of the logic of the driving circuitry.

This feature is especially useful in the dot matrix L.E.D. indicators which employ a 5×7 array of separate emitters with only five plus seven separate connections, five column wires and seven row wires (Fig. 5.4).

This type of indicator can display all letters and symbols as well as the numerals and in a practical system each dot of the matrix is controlled by a complex decoder called a "Read Only Memory" or R.O.M. The elements of the matrix are addressed either a row at a time or a column at a time, at a rate fast enough to eliminate flicker, a particular L.E.D. being "ON" if both its row connection and its column connection are switched on simultaneously. The same system could be used with filament lamps, but a moment's reflection reveals that each lamp in the matrix would require a diode in series with it to isolate it from the other lamps; with L.E.D.s the diode is inherent in their construction, resulting in a very simple array.

IMPROVED FORMAT

The seven segment format is very easy to fabricate using L.E.D. technology but it does offer rather stylised numerals which can be ambiguous under difficult reading conditions. The reason for this possible ambiguity is the lack of redundancy in the simple format, all the bars used for a particular figure being absolutely vital for accurate interpretation.

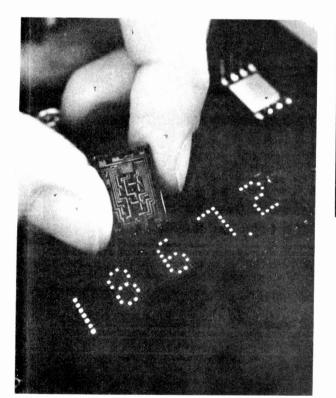
A classic example of this lack of redundancy causing trouble can be envisaged when, for example the readout of a digital-voltmeter has an indicated reading in which the last digit is alternating rapidly between a four and a five, due perhaps to noise on the signal.

With discrete character displays (e.g. cold cathode tubes) this alternation can be easily detected and an accurate reading deduced, but with the seven segment indicator the four and five can be integrated by the observer's eyes to give an apparent reading of nine, which is too far from the correct reading for comfort. Even worse errors can be produced if for any reason one of the bars of the format becomes permanently illuminated or extinguished, as the result of a fault.

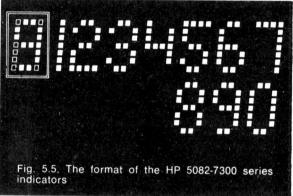
The dot matrix format on the other hand has a great deal of redundancy built in, and even with several dots out of action a readable display is still possible. Unfortunately dot marix displays are both expensive and unnecessarily difficult to drive when only a numerical readout is required. As an answer to this problem a new format has been used in the Hewlett Packard HP 5082-7300 series of indicators, which is really a compromise between a dot and a bar matrix.

As can be seen in Fig. 5.5 the display is still based on a "square 8" format, but in this indicator each bar is made up of dots which can be separately illuminated to give a highly readable and pleasing display with the minimum of complexity. These indicators are unique in also having the decoder/ driver i.c. chip inside the package, only four line B.C.D. inputs being presented to the indicator from the external circuitry.

The actual decoder used in the package will of course be more complicated than a simple seven segment type, but even so the savings involved compared with a 5×7 matrix full alpha-numeric indicator are considerable.



This photograph shows the large size and good readability of the Hewlett-Packard HP5082-7300 series indicators



PRACTICAL DEVICES

The H.P. 5082-730 series of indicators cost in the region of £8 at the moment, which is rather too expensive for amateur use, and this price barrier is also present with the 5×7 alpha-numerics which are currently priced at about £10 each (without decoder). This does not mean that L.E.D. displays are completely excluded from amateur projects because several types of seven segment L.E.D. devices are available at much lower unit cost, and the slight extra cost of even these examples over, say, an incandescent device, can sometimes be justified by the better performance of the L.E.D. types.

Some examples of L.E.D. indicators suitable for amateur use are shown in the photographs.

Next month: Practical display systems using L.E.D. indicators.



Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

RHYTHM BOXES

A custom designed MOS nitride integrated circuit has been designed by **SGS/ATES** for use in rhythm boxes in electronic organs.

The design was developed especially for Eminent and Solina of Holland, makers of high class organs. The research department of Eminent and a professional drummer looked after the musical part. For the electronic part the Eminent engineers worked closely with the SGS MOS design team in Italy.

The resulting rhythm boxes are sold under the trade name "Rithmix" either as built-in optional accessories or as independent units.

The MOS rhythm generator consists of eight different pattern configurations selectable externally by simple pushbuttons. Each pattern can trigger the reproduction of the sounds of up to 12 different percussion instruments. complete musical simulations, also reduces costs and the overall size of the unit.

The range of rhythm boxes available includes, in addition to the simple foot-operated start-stop types, a version presenting an even more far reaching innovation whereby the speed of the rhythm automatically follows the speed of the player.

PIEZO ELECTRIC LIGHT

Some time ago we published a report from Europe in which the reporter mentioned a mains operated gas lighter installed in most kitchens in France. This report created a considerable amount of interest and we are still receiving letters a year later.

Ideal for the Home, Industry and Camping, we have now received details of a new piezo gas igniter from Germany which is guaranteed for ten years, does not use any flints, requires no battery and is not mains operated.

Called the Junkers Piezo Gas Igniter, the "gun" works on the principle that if a sudden force is applied at the interfaces of various crystals, an electric charge is formed. This charge is sufficient to ignite all fuel gases, including natural gas.

When the spring loaded trigger is pulled it releases an impact pin which deforms the piezo crystal No more worries about whether the matches have been left where young children can get at them and cause havoc?

OPEN TO THE

The do-it-yourself electronic enthusiast and the small user of electronic components can now take advantage of the large range of components stocked by **GSPK** (Sales) Ltd., which have only until now, been available to the trade.

They have now issued a new catalogue containing a full range of components available through mail order or from their premises at Harrogate. The price of the catalogue is 10p to cover postage.

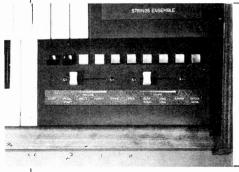
Copies of the catalogue can be obtained from GSPK (Sales) Ltd., Hookstone Park, Harrogate, Yorkshire HG2 7BU.

DISPLAY MODULES

The Codicount range of counting, storage and digital display modules from **Contraves Ltd.**, has been produced to meet the broad requirements of general purpose counting.

The range is claimed to provide designers of control consoles and panels with a "building brick" system which relieves them of logic design responsibility, facilitates rapid updating during development and does not require rack mounted hardware.

The system provides ten variations



Rhythm unit designed by SGS/ ATES

The sounds which they generate form an accompaniment for the organist, providing him with such rhythms as march, swing, rock, slow rock, cha-cha-cha, samba, bossa nova, or with a combination of these. A counter, integrated in the device, driven by an external adjustable oscillator, can sequentially scan the selected pattern to generate the various rhythms.

Although the idea of rhythm accompaniment is not new, the novelty of the current product lies in the use of the most advanced semiconductor technology, which, while it offers improved and more Codicount storing and display module

and causes a voltage discharge of approximately 20kV to appear at the end of the "gun barrel". The end of the "barrel" can be likened to the car spark plug and the voltage appearing at the centre electrode jumps the gap to the barrel wall. It is this 20kV spark, having a pulse duration of only 50 microseconds, that ignites the gas.

The price of the Junkers Gas Igniter seems to depend where the purchase is made, and varies from £2 to £2.40 each. The one we tried was obtained from **Servitronix Ltd.**, 572 Kingston Road, Raynes Park, SW20 8DR; price £2.25.



Junkers Piezo Gas Igniter

and is TTL compatible. Employing i.c.s, the modules have good frequency response, high reliability and freedom from noise problems, the +5V logic supply terminal on each module being decoupled from line noise by a tantalum capacitor.

The supply voltages required for the Logic System is +5V at 21mA to 105mA and +250V at 2.2mA for the readout display.

Further details of the Codicount range of counting, storing and display modules can be obtained from Contraves Industrial Products Ltd., Time House, Station Approach, Ruislip, Middlesex HA4 8LH.





Haematoma associated with tumour



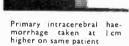
Haematoma

248.

1B



Primary intracerebral haemorrhage



These four photographs show the capability of the new system. They help the diagnostician identify conditions impossible to see using conventional techniques

Computerised X-ray System Improves Pictures of Brain

Diagnosis of brain disease is one of the most challenging aspects of medicine, so this new system from EMI, which makes possible study of brain tissue in detail never before attainable, must be regarded as a major step forward.

This British breakthrough makes use of a computer in processing data from a highly sensitive X-ray machine, and follows three years of research and development by EMI together with specialists from the Department of Health and Social Security.

The system, known as Computerised Transverse Axial Tomography, uses an X-ray machine which rotates around the head of the patient taking over 56.000 readings from a narrow beam of X-rays passing through the head in a single plane. These readings are stored by the computer which then solves 28.000 simultaneous equations to build up a 160×160 matrix picture. For general clinical work an 80×80 matrix has proved adequate.

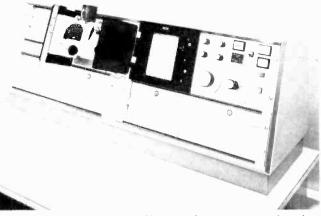
work an 80×80 matrix has proved adequate. The system may be used for investigating a wide range of conditions such as tumours, cysts and haemorrhages. It is claimed to be 100 times more sensitive than conventional X-ray techniques.

The equipment can be handled by one operator with only a modicum of medical knowledge and can screen up to four patients an hour. This speed together with the lack of discomfort to the patient makes the system ideal for use on outpatients who would normally take up valuable hospital beds and staff.

The limitations of conventional X-ray techniques are due primarily to low sensitivity which means that a feature has to be comparatively thick or dense to register on the photograph at all. The fact that X-ray plates record three-dimensional information on a two-dimensional surface also causes difficulties.

The unit as it is used. The cathode ray tube viewing unit and computer are located nearby





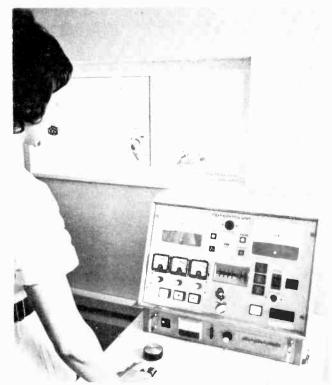
The results from the X-ray equipment are produced as a detailed picture on this cathode ray tube viewing unit

Studies of conventional X-ray techniques have shown that only one hundredth of the information potentially available from the radiation of X-rays is actually realised on the photographic plate.

The difficulties mentioned above are compounded in brain examination since the dense bone tissue of the skull completely surrounds the soft tissue of the brain so that variations in bone thickness completely obliterates details of the brain tissue.

This patented equipment has been under evaluation at Atkinson Morley Hospital, Wimbledon during the last six months. The Department of Health and Social Security is considering installing two of these machines at other hospitals in the near future.

This is the control unit of the installation at the Atkinson Morley Hospital



NEWS BRIEFS

The Open University

A NEW development in the Open University's activities is the provision of Post Experience Courses. These courses are designed for people in all kinds of employment who would benefit from further education or training but who cannot be released from work to take conventional courses.

As with undergraduate courses, tuition will be through correspondence material, supplemented by television and radio programmes, use of computer terminals, home experiment kits, tutorials, and summer schools. Courses will generally run for one year or less, and start January 1973.

Open University certificates will be awarded on completion of the courses.

The first five courses are: Biological Bases of Behaviour, Computing and Computers, Electromagnetics and Electronics, Reading Development, and Reformation Studies.

Electromagnetics and electronics

The aim of the course is to provide a good understanding of the scientific bases of electronic circuit design. It is intended for people in industry, government establishments, hospitals, etc., who find they have a need for electronics, even if they do not intend to proceed to higher level courses. Little prior knowledge of electronics or electromagnetics is required, but the course does assume a background of scientific or technical education beyond GCE "O" level.

A good deal of electronic experimentation at home will be involved, centred round a cathode ray oscilloscope, a signal generator and other apparatus and components loaned to students.

Copies of the Post Experience Courses prospectus and application forms may be obtained from The Post-Experience Student Office, The Open University, P.O. Box 76. Bletchley, Bucks.

New Course for Teachers

A NEW main course in engineering science and technology is planned by Edge Hill College of Education, St. Helens Road, Ormskirk, Lancs. Entitled Science and People, this course is aimed at providing intending teachers with a knowledge of the principles and practice of modern technology.

It is hoped it will also appeal to suitably qualified men and women in industry who wish to enter the teaching profession.

The three year course commences in September, 1972 and leads to the Certificate of Education. Details of admission procedure are available from the Admissions Secretary.

British Amateur Electronics Club

T HE seventh annual B.A.E.C. exhibition will be held on July 22 to July 29 next at the Shelter in the centre of the Esplanade, Penarth, Glamorgan at 7 p.m. every night. This year the club is exhibiting various projects made by members which should make this exhibition more interesting than before.



ART-IN A WORD

Difficult though it may be to accept, we are probably all sensation seekers of one kind or another. In the old times people were wont to get their kicks from the consultation of oracles and inspection within the depths of such things as crystal balls or tea leaves. To some degree, anyway, the random nature of such aids almost certainly held the most valid reason for their success; indeed, this is likely to be the secret of their attraction even now.

However, today we witness the availability of more varied and "way-out" sensations plus an attendant increase in would-be participants to appreciate them. Remember the demand, a couple of years ago, for those quaint little blocks of glass-like plastics containing a number of (seemingly, lit-from-theair) randomly flashing neon lamps? And how about those ever popular collections of Victoriana composed of countless bits of glass suspended from the ceiling and gently tinkling a coolness into the still air of a hot summer evening?

Undoubtedly, the ready-made interest factor is that unpredictable ingredient inherent in randomly produced material or events, one, in fact, which is currently providing the basis for many new art-forms.

One area, in particular, specialises in the production of "unique" forms of both music and prose, relying on computer programs to mould random material according to formalised laws for the specific discipline. This is a field of programming 1 must confess to indulging in myself, and one which it is tempting to replicate, in part at least, in the form of real-time hardware.

In case you have a yen to build something of this nature, Fig. 1., shows just the "bare bones" of a scheme designed to generate random words. It doesn't, of course, include (as a program might do) rules for ordering the occurrence of consonants or vowels, and, indeed, harsh economy dictates a total word length of only four letters.

The hardware model suggested employs one white-noise generator per letter channel which is

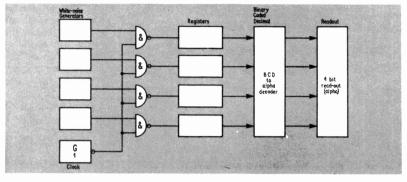


Fig. 1

periodically gated, by a clock pulse. into a register having a maximum count of 26 for the letters of the alphabet. The count: i.e., letter, held in each register, following a clockpulse, is decoded into a form suitable for driving a 16-segment filament type readout and subsequently displayed.

This, possibly crass, waste of electronics may not be your particular idea of a "hang-up," but then it is fairly innocuous and (if you take the trouble to jot down every word) might just write you a sonnet or two!



PICTURE IF YOU WILL

On a different tack, but interesting none-the-less, is a clever new device just given birth by some workers in Albuquerque, U.S.A. It constitutes an unusual way of storing images, and functions in a fashion similar to that of liquid crystal devices employing the nematic phase.

The main difference is that it doesn't use liquid crystals. The prototype device consists of a thin slice of electro-optic ceramic plate coated (on one side) with a photoconductive film with the two sandwiched between a couple of transparent electrodes.

Prior to use, the device must be initially exposed to light (from the film side) while a potential is applied between the electrodes. If, following this, the image from a photographic negative is projected on to the film side, and the voltage applied once again, a latent image will be stored by the ceramic.

This image can be subsequently reproduced simply by illuminating the device from the film side; the electrodes at such times must not be energised, otherwise the image could be erased. Actual erasure takes place by re-exposing to light while the voltage is applied, and in this way continues countless images can be stored again and again.

The principle Cerampic, as the device is called, employs for its operation is one which relies upon the formation of ferro-magnetic domains having different orientations within the ceramic plate.

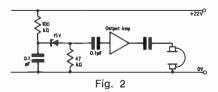
Variation in the orientation constitutes the difference in the areas of contrast in the stored image. This, seemingly, controls the amount of light scattering produced by the material; thus, bright parts of the image seem to be formed by domain orientations aligned with the direction of incident light, while darker areas are produced by alignments which scatter more of the light away.

Since this remarkable material can be switched on and off at fairly high rates, it would appear, like liquid crystals, to be a serious contender in the race to produce picture-frame television. Even if it never gets that far, it is sure to provide sufficiently good resolution for such things as document transmission and copying by wire.

NO \$HHH!!

In the last On the Fringe under the heading "Occultaphonics" I mentioned a device which could be put together to produce white-noise. Unfortunately, the circuit, shown in the accompanying figure at that time, did not include a Zener diode. This component, vital to the operation of the device, is now shown in Fig. 2.

Some Zeners are a bit reluctant to generate electrical "hash," so you may well need to try a few before finding the ideal.





There are times when it is Impossible to answer a call at the door. The "Callercord" will answer callers and record messages on magnetic tape.



THERE are times when it is inconvenient or impossible to answer a call at the front door. You could be in the bath, taking a nap, or out shopping with no one else around to answer the door. The Callercord has been designed to help alleviate the frustrations caused to you or the caller in these and other situations by enabling you to leave a recorded message which is automatically played back to the callers each time the door bell is rung. Having repeated its message, the machine then records any message which they may have for you. The Callercord could be of particular use to small shopkeepers who could use it to take messages or orders while the shop is closed, as for example during lunch hours or on early closing days.

The Callercord uses two tape recorders, but in conjunction with just one recorder, it can easily provide other services, such as repeatable information or commentary at any place where it is not necessarily desirable to have the information constantly repeating itself and where it need only be available at the touch of a switch.

Again using one tape recorder, preferably a larger one than those used in the Callercord, and with slight modification to the sensing device, the machine can even be used as a baby-soother, sensing when the baby wakes and playing soothing music or mother's voice.

Without any tape recorders at all, the main control circuit can be used on its own to control a variety of functions for which automatic sequential timing is required, such as remote control of photographic lighting, cameras and enlargers. The individual sub-units of the control can also be duplicated to control an electric train or model car.

DESIGN CONSIDERATIONS

The basic design considerations are for the Callercord to sense when the door bell is being rung, play back a pre-recorded message (PBM), then record a message from the caller (RCM), and finally switch off the recorders.

There are several ways of achieving all these functions and the first consideration is the method to be used for record and playback. It is possible to use an automatic record player for PBM, but unless the message stays unchanged, the cost and inconvenience of having discs cut rules out this method. It is also feasible to use magnetic disc recorders but these are hard to come by.

Another method, and one which at first sight appears attractive, is to use a stereo tape recorder, but the problem here, as will be seen later, is that of repeatedly playing the PBM. Cassette recorders were seriously thought about for a while, but the cheapest that could be found was about £13. It was decided that two cheap tape recorders were the best answer, and a search was started to find the cheapest ones possible. A number were found, and the make finally chosen was the Japanese "Parrot Bookcorder". The quality of sound on these might not suit some hi-fi enthusiasts, but in this instance, sound quality was considered to be unimportant, and these machines have proved to be satisfactory for the job.

BLOCK DIAGRAM

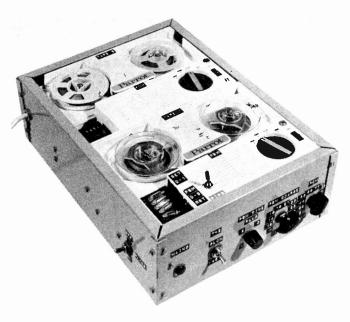
Sensing of the door bell ringing could be achieved by placing a microphone so that it picks up the bell, and then passes this signal as a start pulse to the control circuit. In this machine a similar effect is achieved by breaking connection with the input part of the control circuit when the bell switch is used. To keep the number of wires from the door to a minimum, the connection between the input control and the door is made using the same wires that feed the external speaker. This can be seen in the block diagram of Fig. 1.

The bell switch S1 is a push to make and break type with one side in series with the speaker leads. When connection with the control circuit input is broken, it passes a pulse through to a bistable gating circuit. This only allows one pulse through and then closes to prevent further pulses from disrupting the rest of the circuit, and will not open again until the control circuit has completed its cycle.

As the gate closes, a pulse initiates a delay (Delay 1) before anything else happens, so that there is time for someone to answer the door personally if they wish to do so. At the end of the delay, two relays are operated. Relay RLA switches the external speaker LS1 out of circuit with the control input and across to relay RLB which switches the speaker leads into circuit with the PBM recorder (Tape Recorder 1) and also causes it to play back its pre-recorded message.

AUTOMATIC REWIND

There is a choice of ways to make the PBM recorder play back the same message each time it is needed. It is of course possible to record the same message over and over again onto a whole spool of tape, but the prospect of doing this each time the message has to be changed is a little daunting, and there is also the problem of the control circuit selecting the right moment at which to switch on the RCM recorder (Tape Recorder 2). Another method is to use an endless loop of tape, switching over to the RCM recorder at a point determined either by



time or by a physical method such as using a contact switch or a photoelectric sensor. The endless loop system was in fact experimented with, but was eventually rejected because of the problem of ensuring reliable loop transport plus the fact that the "Parrot" has a spool drive which would need adapting to pull the tape across the heads.

The method finally chosen uses a spool of tape which is rewound automatically after playing the message, thus allowing for repetition and length variation as required. Automatic rewinding initially presented another problem, for the "Parrot" has a

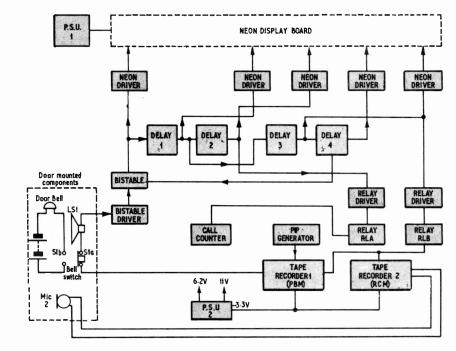


Fig. 1. Block diagram of Callercord

COMPONENTS . . .

Resisto		D (0	0.01.0
R1	47kΩ	R46 R47	8·2kΩ
R2 R3	4·7kΩ 4·7kΩ	R47 R48	8·2kΩ 8·2kΩ
R4	3·3kΩ	R40 R49	8·2kΩ
R5	8·2kΩ	R50	8·2kΩ
R6	8·2kΩ	R50	2·2k Ω
R7	3·3kΩ	R52	10kΩ
R8	4·7kΩ	R53	220 \
R9	10k Ω	R54	47kΩ
R10	1kΩ	R55	1kΩ
R11	1kΩ	R56	10kΩ
R12	10k Ω	R57	10kΩ
R13	18k Ω	R58	1kΩ
R14	2·2k Ω	R59	1kΩ
R15	10k Ω	R60	18k Ω
R16	220 Ω	R61	10k Ω
R17	10k Ω	R62	2·2kΩ
R18	18k Ω	R63	10k Ω
R19	1 2 0kΩ	R64	220 Ω
R20	1kΩ	R65	68kΩ
R21	22kΩ	R66	1kΩ
R22	22kΩ	R67	18kΩ
R23	22kΩ	R68	10kΩ
R24	22kΩ	R69	120kΩ
R25	22kΩ	R70	220kΩ
R26	22kΩ	R71	680Ω 1W
R27	22kΩ 10kΩ	R72 R73	5·6k Ω 270Ω
R28 R29	1kΩ	R74	470Ω
R30	1kΩ	R75	68Ω 1W
R31	2·2kΩ	R76	12Ω 2 1 W
R32	10k Ω	R77	33Ω 1W
R33	220 Ω	R78	33Ω 1W
R34	47kΩ	R79	120kΩ
R35	1kΩ	R80	22Ω 1W
R36	18kΩ	R81	4·7kΩ
R37	10kΩ	R82	220kΩ
R38	10k Ω	R83	220k Ω
R39	10k Ω	R84	4·7kΩ
R40	1kΩ	R85	4·7kΩ
R41	1kΩ	R86	27kΩ
R42	8·2kΩ	R87	27k Ω
R43	8·2k Ω	R88	10kΩ
R44	8·2kΩ	R89	22 Ω
R45	8·2kΩ	R90	4·7k Ω
	0%, ∔W carbon		
Capacit	ors		
Ċ1	0·1µF		
C2	0.01µF		
C3	0·01µF		
C4	47µF elect 6V		
C5	0·01µF		
C6	470µF elect 6V		
C7	0.01		
C8	320µF elect 6V		
C9	0.01µF		
C10	10µF elect 6V		
C11 C12	0·01µF 32µF elect 450V		
C12 C13	32μ F elect 450V 32μ F elect 450V		
C13	2,000µF elect 25V		
C14	470µF elect 6V		
C15	100µF elect 12V		
C17	470 F elect 6V		

Potentiometers 10kΩ carbon linear VR1 2.5kΩ carbon linear VR2

Transist	ors			
TR1	BC109	TR14	BC109	
TR2	BC109	TR15	BC109	
TR3	BC109	TR16	BC109	
		TR17	2N2102	
TR4	BC109			
TR5	2N2102	TR18	BC109	
TR6	BC109	TR19	BC109	
TR7	BC109	TR20	BC109	
TR8	2N2102	TR21	BC109	
TR9	BC109	TR22	2N2102	
TR10	BC109	TR23	BC109	
TR11	BC109	TR24	BC109	
TR12	2N2102	TR25	2N3055	
TR13	BC109	TR26	R30 BC109 (5 off)

Diodes

IN914 (6 off)
OA91 (2 off)
BY100
BYX22/200 (4 off)
BZX61/C6V2
BZY88/C6V2
OA91
BZX61/C5V6
IN914 (3 off)
BZY88/C3V3
OA91 (3 off)

Relays

RLĂ/RLB 6-12V, 185 ohms, 4 sets of changeover contacts, type PC4 (2 off)

Transformer

230V Primary, 6.3V + 6.3V secondary. T1 Hygrade Filament transformer

Solenoid

12V Solenoid, thrust type operation X1

Counter

12V, four figure electromagnetic counter X2

Switches

- S1 Press to make and break
- Š2 Single pole 8-way rotary switch **S**3
- Single pole 10-way rotary switch D.P.D.T. mains toggle Š4
- S.P.S.T. toggle **S**5
- **S**6 S.P.S.T. toggle

Miscellaneous

LP1-LP5 Miniature 90V neons (5 off) LP6 Mains indicator neon "Parrot" Book-Corder tape recorders (2 off) $4\frac{3}{2}$ in × $3\frac{1}{2}$ in, 0-1in matrix Veroboards (2 off) $4\frac{3}{4}$ in $\times 3\frac{1}{2}$ in, 0.15in matrix Veroboard (1 off) FS1-250mA antisurge fuse Universal chassis 13in × 10in × 4in $\frac{1}{2}$ in $\times \frac{1}{2}$ in angle aluminium, 13in long (2 off) 8-way edge connector sockets and plugs (5 off) Octal plug and socket

0.68µF

0.01µF

0.01µF

0.01µF

470μF elect 6V 470μF elect 6V

470µF elect 6V 470μF elect 6V 0·68μF

C17

C18 C19

C20 C21

C22 Č23

C24

C25

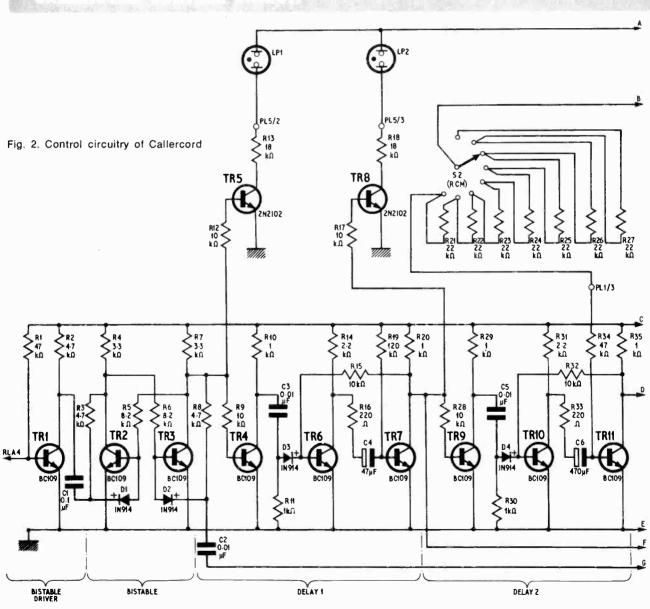
mechanical rewind control. This works by rocking a double-ended motor on a pivot, one spindle driving forwards, and after rocking over, the other driving backwards. By slightly modifying the motor bracket and attaching a solenoid, automatic wind and rewind were easily achieved.

With the solenoid switched on, Recorder 1 drives forward, and backward when off. The supply line is switched over to relay RLB from RLA and is the power line to activate the solenoid. Thus when relay RLA is on, RLB has control of the direction of Recorder 1. To prevent the tape message being heard in reverse, RLB also switches off the loudspeaker during rewind.

The point at which Recorder I switches into reverse is determined by another delay circuit (Delay 3) preset by an external switch to suit the length of message and initiated by the ending of Delay I. At the end of Delay 3, RLB is turned off, so switching Recorder 1 to rewind, and Recorder 2 to record the incoming message, and at the same time switching off the external speaker.

ADDITIONAL DELAYS

Ideally the machine should sense the right moment at which to switch off Recorder 2. This could be done by sensing the pause after the caller has finished speaking, but it could be subject to error, either because of the caller hesitating too long between sentences, or because of possible noises from the street. Consequently, switching off is achieved by again using an externally preset time delay (Delay 2) which is usually preset to give about 40 seconds of recording time, though it can be set to give other times if necessary. For the sake of con-



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venience this delay is also started by the ending of Delay 1.

At the end of Delay 2, several things happen. Relay RLA is switched off, so switching off the power to the recorders and solenoid. Although Recorder 1 has by this time rewound the tape, it still has its motor on. While the motor is unable to turn and is drawing maximum current, it will come to no harm during the short time that this condition prevails. Relay changeover also means that the speaker leads are also switched back into circuit with the control input, and the call counter, which clocks up incoming calls, is disconnected.

The ending of Delay 3 initiates a fourth delay circuit (Delay 4) which allows a short pause and then sends the open signal to the gate. The cycle is now complete, and the circuit is ready to receive the next caller.

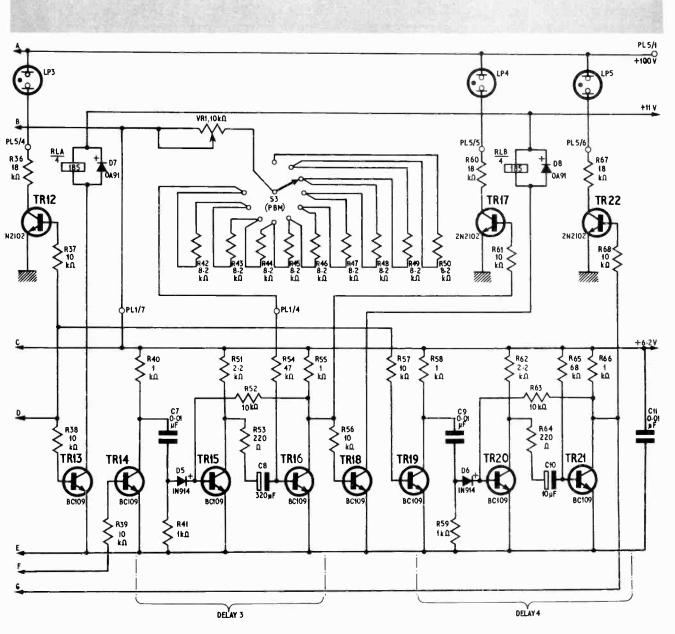
NEON INDICATORS

All of the various processes in the control chain are indicated by five neon lamps. Besides providing a visual check-out for correct functioning of the stages, they facilitate delay time settings.

CONTROL CIRCUIT ACTION

The start pulse is derived in a very simple manner as seen in the control circuit of Fig. 2. The base of TR1 is taken to earth via the leads of the external speaker, so holding TR1 off. In series with the speaker is placed a switch which is normally closed. The same press switch has two normally off contacts which are in series with the door bell. On pressing the switch, the bell is rung and the base of TR1 is disconnected from earth.

When this happens, the base of TR1 is made positive through R1 and a negative going collector



voltage is passed as a pulse via C1 to the bistable circuit, so triggering it into its opposite state. The negative going voltage change of TR3 is inverted by TR4 and this drives the monostable circuit (Delay 1) into its quasistable state with TR7 collector negative going. The voltage change is inverted by TR9 and is passed as a pulse to trigger the monostable, Delay 2. The transistor TR13 then conducts thus turning on relay RLA.

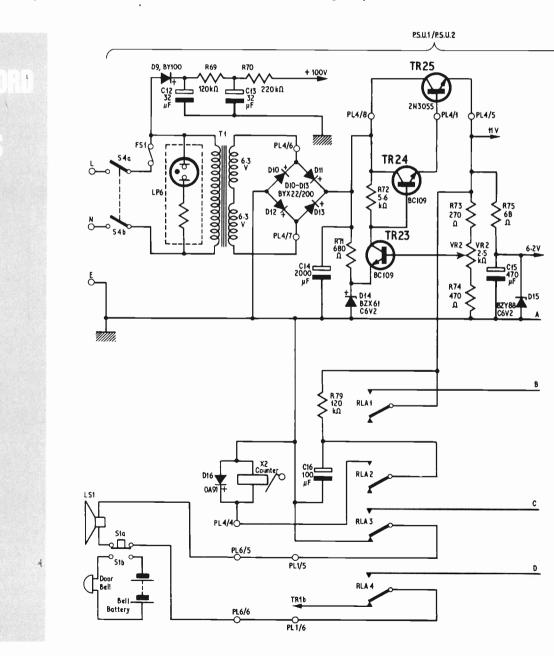
The voltage change of TR7 is also inverted by TR14 and passed as a pulse to trigger Delay 3 monostable, so turning on relay RLB. When TR16 reverts to its stable state, TR18 ceases to conduct; this relay switches off. When TR11 reverts to its stable state, TR13 ceases to conduct, turning off relay RLA. This second voltage change of TR11 is inverted by TR19 and triggers Delay 4 As TR21 stabilises, it passes a pulse to TR3 reverting the bistable gate to its original state. The control circuit is then receptive to the next start pulse.

VARYING THE DELAY TIME

The switching time of the delay circuits 2 and 3, may be changed by controlling the charge rate of the respective timing capacitor by switching in resistors. With Delay 3 fine control is achieved by using a variable resistor in series with the switch S3.

PULSE SUPPRESSION

The diodes that are used with the relays, solenoid, and recorder motors are vital to the smooth functioning of the control circuits, for at the moment of switching off a coil, a back e.m.f. is developed which by feeding back along the power line could influence



other circuits, and in the case of the monostables and bistable, could be large enough to trigger them into their opposite states.

POWER SUPPLIES

The main power supply (Fig. 3) has the transformer secondary windings connected to give 12.6V a.c. which is rectified and smoothed, and the resulting d.c. supplies the stabilising circuit which has its output voltage set at 11V. The 11V supplies the relays, solenoid and counter, and is also sub-divided and dropped to give 6.2V for the transistor circuits, -3.3V for the tape amplifiers, -3.3V for the motor of Recorder 1, and as a convenience, a switchable $-1\frac{1}{2}V/-4V$ to allow Recorder 2 to move slowly for recording, and fast for rewinding.

Something not foreseen was the fact that the nomi-

nal rectified voltage of about 18V with no load, drops down to about 13.5V under full load. This is due mainly to distortions occurring in the a.c. waveform from the transformer secondary windings, and partly due to the impedance of the transformer and rectifiers. As the stabilising circuit needs a voltage drop of at least 2V across it to give adequate stabilising, the output voltage has been set at 11V to allow for the full load condition. If a 15V transformer had been available, then the stabilised voltage would have been set at 12V. However, this is not an important point as the solenoid does not need to have an accurately controlled supply, and the 11V fed to the 12V relays is still within their operating range of 8V to 17V.

Next month: Full constructional details for the Callercord.

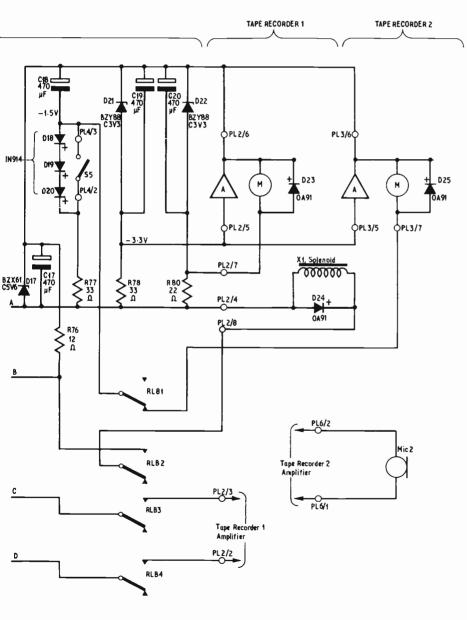


Fig. 3. Power supplies and relay contact inter-wiring

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PATENTS REVIEW...

IMPROVING POWER AMPLIFIERS

TRANSISTOR power amplifiers for broadcast transmitters normally use output stages with a pair of transistors working in "pushpull". The transistors can be biased into class A, AB or B and the type of bandwidth required is usually 1.5MHz to 30MHz.

In British Patent 1 257 550 the Marconi Company Ltd. explain how the design requirements for a transformer to couple balanced transistors to an unbalanced aerial are mutually conflicting. Some factors dictate a large transformer with a large number of turns while other factors dictate a small transformer with a small number of turns.

The Marconi patent is directed to a transmitter power amplifier with an auto-transformer having a centre tapped primary winding. The output transistors are connected in push-pull to feed the centre tap and in addition a conventional transformer is used, the secondary winding of the autotransformer being connected to the primary of this conventional transformer and the output taken from its secondary.

By using an auto-transformer a much lower leakage inductance may be achieved than with a conventional transformer for the same power handling requirements. Closer matching of the output impedance of the transistors may

BP 1 257 550 fig. 1a

Fig. 1b

also be obtained. The use of a conventional transformer provides the necessary function of impedance transference and conversion from a balanced to an unbalanced load.

Fig. 1a shows such a circuit compared with its equivalent in Fig. 1b. The power transistors TR1 and TR2 feed the centre-tapped auto-transformer T1 which in turn feeds the conventional transformer T2 coupled to the aerial. The capacitances of C2, C3 and C4 are, in part, formed by stray capacitances and in part by capacitors.

As the equivalent circuit Fig. 1b shows (L1. 2 and L3. 4 being the leakage inductances of the autotransformer and the conventional respectively), transformer the arrangement functions as a lowpass filter with a top cut-off equivalent to the highest frequency transmitted. The capacitances and inductances require careful selection but the overall advantages can be low leakage despite high power handling and good conversion from balanced to unbalanced load. There will also be tight coupling between the push and pull sides of the windings and isolation of the load from the transistor supply voltage.

NEON NOVELTY

P LANS for an attractive recreational device of the type that is sometimes written off as "a gift for someone who has everything" are to be found in British Patent 1 257 288, in the name of Roger Saunders.

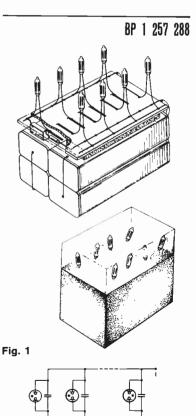
The idea is to envelope neon lamps in a block of transparent plastics with the circuitry for periodically illuminating them in a block of opaque plastics. See Fig. 1.

It is perhaps not generally recognised that if sufficient neon lamps are flashed in a complicated sequence, the psychological effect can be that the flashing is nonperiodic and non-repetitive. By hiding the circuitry and batteries in opaque plastics, the layman is even more confused.

Saunders illustrates eight neon lamps and four cells buried in acrylics. In the simple circuit diagram given (see Fig. 2), the neons are shown all paralleled across the battery of cells. Each neon is bridged by a capacitor and the pairs of paralleled neons and capacitors are each in series with a resistor. The neons can be divided into groups and, for instance, in one group be connected across capacitors of one value (0.22μ F) and in the other group be connected across capacitors of another value (0.47μ F). In each of these two groups different value resistors are used (e.g. 5.6, 7.5, 8.2 and 10 megohms).

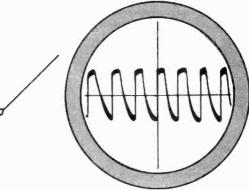
Where the triggering voltage of the neons is 86 volts the inventor suggests a battery voltage of 90 volts. Battery life may well be several years with the low current requirements of the neons.

Certainly not a world-shattering technical breakthrough, but a worthwhile addition to the currently fashionable range of useless but attractive functional objects.





electronics really mastered



2/ READ, DRAW

AND UNDERSTAND

CIRCUIT DIAGRAMS

as used currently in the

various fields of electronics.

step by step, we take you through all the fundamentals of electronics and show how easily the subject can be mastered. Write for the free brochure now which explains our system.

BUILD. SEE AND LEARN

1/ BUILD AN

electronics.

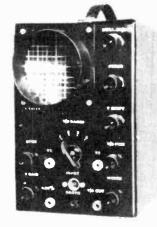
OSCILLOSCOPE

You learn how to build an oscilloscope which remains your property. With it, you will become familiar with all the

components used in

... practical ... visual ... exciting !

no previous knowledge no unnecessary theory no ''maths''



RAPY

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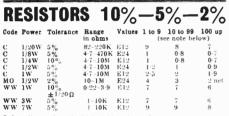
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*AD162	PNP	Med. power	36
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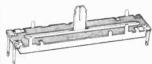
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470/3, 6p each. 470/3, 8p each. 47/50, 47/63, 100/35, 470/10, 7p each: 100/50, 220/35, 9p each: 100/83, 470/23, 100/10, 10p each: 220/63, 470/35, 1000/16, 14p each: 1000/23, 18p each: 220/63, 1000/33, 10p each: 2200/25, 30p each: 1000/63, 2200/35, 4700/16, 33p each.

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TRANSELECTOR

LOGIC Semiconductor .identifier

By T. P. SCHOFIELD

PROBABLY cne of the cheapest ways for an amateur to obtain transistors is to buy a pack of "unmarked, untested transistors" and test them himself. This is useful for experimental work in which a large number of transistors is required which need not conform to a particular stringent specification.

This automatic transistor selector is designed to test batches of such transistors quickly and simply; it also tests diodes, since an inoperative transistor can frequently be used as an excellent diode.

The device to be tested is plugged into a transistor holder; the selector indicates by lighting lamps whether it is a *pnp*, *npn*, or a diode with anode at collector socket, or diode with anode at emitter socket. The logic functions required to ascertain these properties are performed by TTL logic i.c.s in the SN74N series.

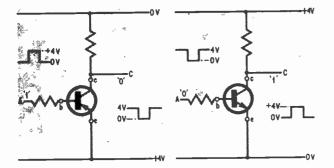


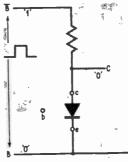
Fig. 1a. A logic "1" to the base and a "0" at collector will indicate a pnp transistor when emitter is at +4V

Fig. 1b. A logic "0" to the base and a "1" at collector will indicate an *npn* transistor when emitter is at 0V It seems possible that with additions to its circuitry the selector could perform more complex tests to separate transistors with a particular frequency response or gain.

TESTING PRINCIPLES

The tests applied are shown in Fig. 1, assuming of course that lead connections are established. Transistors are tested by biasing them as shown in a common-emitter mode logic circuit—and feeding in a test signal. With a *pnp* transistor, if A is at about +4V then C should be driven to about 0V; if A is at about 0V then C should be at about +4V. The same is true of an *npn* transistor in its circuit.

For the diode test, a square wave of alternately 0V and +4V is applied to \overline{B} . A diode in the test socket with its anode in the collector terminal, and



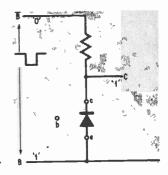
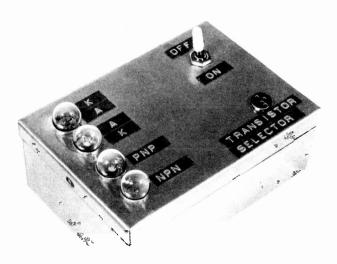


Fig. 1c. Diode polarity (cathode to "e" terminal) when logic 1 fed into B line gives 0 at C Fig. 1d. Diode polarity (anode to "e" terminal) when logic 0 at B gives 1 at C

In both Figs. 1c and 1d, both B and B lines are floating



its cathode in the emitter terminal, will conduct only when \overline{B} is positive relative to B and C will be at 0V. A diode the other way round will conduct only when \overline{B} is negative relative to B and C will be at a steady 4V.

WORKING LOGIC CIRCUIT

The complete circuit diagram of the unit is shown in Fig. 2.

The timebase for the test waveform is derived from a Schmitt triggered multivibrator formed by half of IC1. This circuit, a clock pulse generator, is very reliable, simple to construct using only two external components, and produces an output designed to be compatible with TTL circuits and variable from 10MHz down by changing the value of C1. It is possible to use other oscillators, but they must produce a very square wave or the transistor under test may be triggered differently from the remaining logic circuitry, producing a false result.

The output of the clock pulse generator (labelled A in Fig. 2) is fed to one half of an SN7474N—dual edge-triggered type D flip-flop—arranged to divide by two. The true and inverted outputs of this flip-flop, called B and B in the diagrams, are used to bias the emitter and collector terminals of the transistor under test. The output of the oscillator is supplied as a test signal to the base of this transistor.

As B changes the transistor is biased as shown in Fig. 4 for testing pnp and npn transistors and diodes. Thus the waveform appearing at C contains all the necessary information about the device under test. This is amplified by TR1 so that it is acceptable to TTL inputs. The truth table (Table 1) shows what the voltage (\overline{C}) at the collector of TR1 should be when different devices are tested. The truth table shows the logic states 0 and 1 representing 0V and +4V respectively.

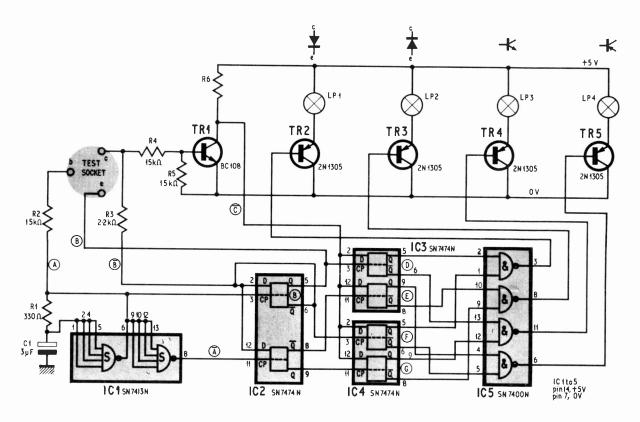


Fig. 2. Circuit diagram of logic transistor and diode identifier





YATES ELECTRO (FLITWICK) LTD. ELSTOW STORAGE D KEMPSTON HARDW BEDFORD	EPOT Catalogue which contains data sheets for most of the components listed will be sent free on request
¹ / ₂ 10% 3·3MΩ−10MΩ E12 1p 0 ¹ / ₂ 2% 10Ω−1MΩ E12 3·5p	160V:0-01μF,0015μF,0022μF,0-033μF,0-047μF,0-068μF,3p.0-1μF 3±p.0-15μF 4±p 160V:0-01μF,0015μF,0-022μF,0-068μF,11p.1-0μF,13p. 22μF,5p.0-33μF,6p.0-47μF,7±p.0-68μF,11p.1-0μF,13p. MULLARD POLYESTER CAPACITORS C280 SERIES 250V P.C. mounting: 0-01μF,0-015μF,0-022μF,3p.0-033μF,0-047μF,0-068μT,11p.1-0μF,13p 15μF,20p.2-2μF,24p,24p.
DEVELOPMENT PACK 0.5 watt 5% Iskra resistors 5 off each value 4.7Ω to IMΩ. E12 pack 325 resistors £2.40, E24 pack 650 resistors £4.70.	MYLAR FILM CAPACITORS 100V 0:001μF, 0:002μF, 0:01μF, 0:02μF, 0:01μF, 0:02μF, 2ip. 0:04μF, 0:05μF, 0:1μF, 3ip. 100pF to 10,000pF, 2p each.
POTENTIOMETERS Carbon track 5k Ω to 2M Ω , log or linear (log $\frac{1}{2}W$, lin $\frac{1}{2}W$). Single, 12p. Dual gang (stereo), 40p. Single D.P. switch 24p.	 ELECTROLYTIC CAPACITORS—MULLARD C426 SERIES 60 each (uF/V) 10/2-5, 40/2-5, 80/2-5, 160/2-5, 30/2-5, 80/4, 32/4, 64/4, 125/4, 250/4 400/4, 6-4/6-4, 25/6, 4, 50/6-4, 100/6-4, 220/6-4, 3120/6-4, 4/10, 16/10, 32/10, 64/10, 125/10 200/10, 2-5/16, 10/16, 20/16, 40/16, 80/16, 125/16, 1-6/25, 64/25, 12-5/25, 25/25, 50/25 80/25, 1/40, 4/40, 8/40, 16/40, 32/40, 50/40, 0-64/64, 2-5/64, 5/64, 10/64, 20/64, 32/64.
SKELETON PRESET POTENTIOMETERS Linear: 100, 250, 500Ω and decades to 5MΩ. Horizontal or vertical P. mounting (0+1 matrix). Sub-miniature 0-1W, 5p each. Miniature 0-25W, 6p each. TRANSISTORS	C. MULLARD C437 SERIES 100/40, 160/25, 250/16, 400/10, 640/6-4, 800/4, 1000/2-5, 9p. 100/64, 160/40, 250/25 400/16, 640/10, 1250/4, 1000/6-4, 1600/2-5, 12p. 160/64, 250/40, 400/2-5, 640/16 2000/4, 1000/10, 1600/6-4, 2500/2-5, 15p. 250/64, 400/40, 640/25, 3200/4, 1000/16 1600/10, 2500/6-4, 4000/2.5, 18p.
AC107 15p BC108 10p BFY51 21p OCP71 40p 2N3703 11 AC126 12p BC109 10p BFY51 22p OCP71 40p 2N3703 11 AC126 12p BC109 10p BFY51 22p OCP71 40p 2N3704 11 AC127 12p BC147 10p BFY56 32p 2N2269 16p 2N3705 12 AC128 15p BC148 13p OC26 45p 2N22646 60p 2N3707 12 AC131 12p BC149 13p OC26 45p 2N22647 9p 2N3707 12	P ELECTROLYTIC CAPACITORS Miniature P.C. mounting 5p each P (μF/V): 10/12, 50/12, 100/12, 200/12, 5/25, 10/25, 25/25, 100/25. 5p each
AC132 12p BC157 13p OC33 45p 2N29260 9p 2N3708 (i AD140 50p BC158 13p OC42 12p 2N29260 9p 2N3708 (i AD161 33p BC159 13p OC42 12p 2N29260 9p 2N3708 (i AD161 33p BC159 13p OC44 12p 2N29260 9p 2N3710 (i AD162 36p BD131 75p OC70 12p 10p 2N3711 (i) AF114 20p BF179 32p OC71 12p 2N3055 60p 2TX302 13 AF115 20p BF194 15p OC72 12p 2N3055 60p 2TX503 14 AF116 20p BF194 15p OC72 12p 2N3702 13p 2TX503 14 BC107 10p BF50 22p OC82D 12p 2N3702 13p 40362 54 <td>Dp VEROBOARD JACK PLUGS AND SOCKETS p 0-1 0-15 Standard screened 18p 2-5mm insulated 8p p 2½ × 53 22p 16p Standard screened 18p 2-5mm insulated 8p p 2½ × 53 24p 24p Standard screened 15p 3-5mm insulated 8p p 3½ × 5 24p 24p Standard socket 15p 2-5mm socket 8p p 3½ × 5 27p 57p 57p 5tereo socket 18p 3-5mm socket 8p p 17 × 31 100p 78p DI.N. PLUGS AND SOCKETS 9p 2p 2p 5m 5m</td>	Dp VEROBOARD JACK PLUGS AND SOCKETS p 0-1 0-15 Standard screened 18p 2-5mm insulated 8p p 2½ × 53 22p 16p Standard screened 18p 2-5mm insulated 8p p 2½ × 53 24p 24p Standard screened 15p 3-5mm insulated 8p p 3½ × 5 24p 24p Standard socket 15p 2-5mm socket 8p p 3½ × 5 27p 57p 57p 5tereo socket 18p 3-5mm socket 8p p 17 × 31 100p 78p DI.N. PLUGS AND SOCKETS 9p 2p 2p 5m 5m
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BZV10 B00V 6A 25p OA90 BZV13 200V 6A 20p OA91 IN4001 50V IA 7p OA202 IN4000 40V IA 8p IN4146	Pkt. 50 pins 20p 20p 9V mains power supply. Same size as PP9 battery. THERMISTORS VA1055S 15p; VA1066S 15p; VA1077 15p; R53 £1-35. PCOMPACT CASSETTES—IN PLASTIC LIBRARY BOX C90 65p CI20 85p.
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SLIDER POTENTIC/METERS 86mm x 9mm x 16mm, length of track 59mm SINGLE 10kQ, 25kQ, 50kQ, 100kQ, log or 1in DVAL GANG 10kQ + 10kQ, etc. log or 1in Knob for above RONT PANEL 18 gauge panel 12in × 4in with slots cut for use with slider pots. Grey or matt black finish, complete with fixings for 4 pots.	2800μF 100V £3.00 HIGH VOLTAGE TUBULAR CAPACITORS—1,000 VOLT 9 0.01μF 10p 0.047μF 13p 0.22μF 20p 9 0.022μF 12p 0.1μF 13p 0.47μF 22p

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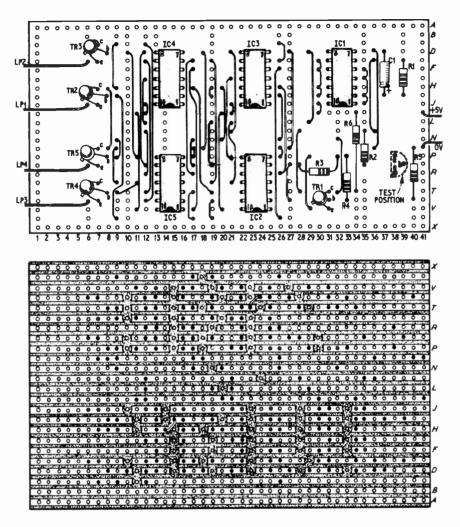


Fig. 3. Suggested circuit layout on Veroboard

COMPONENTS . . .

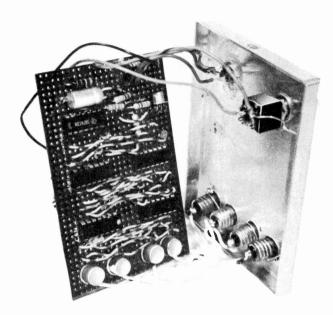
R2	stors 330Ω 15kΩ 2·2kΩ		R5	15kΩ 15kΩ ±10%, ≵ W carbon
TR	sistors I BC108 2 to TR5	2N1305 (4 d	off)	
ICĬ IC2	rated Cir SN7413 to IC4 SN7400	N SN7474N (3	off)	
C1 LP1 B1	ellaneous 3μF elec I to LP4 6V batte roboard 0-	t10V 6V40mA	in >	× 2½in

DECODER

The value of \overline{C} for each state of the testing waveforms A and B are stored in the flip-flops D, E, F, G in IC3 and IC4. (The SN7475N quad flip-flop is not suitable as it is not edge-triggered and its clockinputs are joined in pairs).

When A goes positive, and the value of B changes, either D or F will be clocked; when A goes negative, and the other flip-flop in IC2 follows B, then either E or G will be clocked. This arrangement involves fewer extra i.c.s than one which tries to decode A and B. The inputs of these storage flip-flops are interpreted by ordinary 2-input NAND gates in IC5 according to the truth table. The particular functions chosen make it impossible for the selector to indicate that it is testing a transistor and a diode simultaneously.

The outputs from 1C5 operate lamps indicating the sort of device placed in the test socket, through buffer transistors TR2 to TR5. It may well be found that the SN7400N seems able to light the lamps directly, but is not recommended due to the risk of overloading the gates. Two SN7440N circuits, replacing the SN7400N, could pass the current



without harm if preferred, so making the extra buffers unnecessary.

PRACTICAL LAYOUT

The circuit is easily built on a piece of 0.1in matrix Veroboard, as shown in Fig. 3. This diagram shows the complete layout as seen from the top component side. The copper strips should be broken where indicated. Then those wire links which occur under IC2 and IC4 should be made.

The integrated circuits can be soldered to the board at this stage, and also the resistors.

Next all the link connections shown should be made. It is easiest by far to use single-core, plastic sleeved wire for this. Alternatively the connections could be made with 22 s.w.g. tinned copper wire and separate sleeving.

Finally, the connection wires leading to the battery and bulbs and then the transistors can be soldered in. The prototype was not fitted in a box and the bulbs were consequently just soldered onto the wires, but this is of course a personal preference.

The test socket used, fitted directly onto the board as shown in Fig. 3, but it had four sockets and the two corresponding to the base of a transistor were first soldered together. The socket could be fixed on the outside of a case and joined to the appropriate points on the board by three wires if desired.

POWER SUPPLY

A 6 volt battery seems the ideal power source for the transistor selector, but any adequately smoothed supply of 500mA at 5V would be suitable. The unit takes about 60mA quiescent current so the battery should be disconnected by a switch when not in use.

A Ā	Ā B B	State o Entry for into Transis	r	for Diode				
				flip-flop	npn	pnp	c	e
1	0	0	1	G	1	x	0	1
0	1	0	1	DE	0	X	0	1
1	0	1	0	E	X	1	0	1
0	1	1	0	F	x	0	0	1

It will be noticed that the 6V bulbs are considerably under-run at less than 5V. It was found, however that their indication was still quite clear (since they are always full on or full off) so the provision of 5V bulbs or a separate 6V power supply (which would involve a design change in the output stages) was not thought worthwhile.

TESTING

To test the unit when it is built, the power supply is connected. None of the lamps should stay on (but one might flicker once as the circuits start operating). Check that the appropriate pins of all the i.c.s are connected to the power supply. Place a diode in the test socket; one of the "diode" lamps should light. If it does not, check the operation of the circuit from the multivibrator on. (The multivibrator may be slowed by connecting a much larger capacitor in parallel with Cl.) When the selector has been made to operate properly with a diode, a transistor may be tried on it. Short circuits caused by solder runs are the most likely cause of a fault.

Only one light should come on at a time, and it should not flicker. A flickering lamp indicates a poorly decoupled power supply, an exhausted battery or a faulty connection. Two lamps should not normally both light at the same time. If they do the fault is probably in the wiring supplying IC5. If the wrong lamp lights, one of a number of faults may be responsible and the whole circuit should be checked starting with IC1.

It is probably worthwhile trying out this tester on known transistors and diodes to ascertain correct functioning.

POINTS ARISING

VERSATILE LIGHT EFFECTS UNIT (June 1972)

In Fig. 5 the connections for the base and emitter of TR1 and TR4 should be reversed.

CHARGER POWER UNIT (June 1972)

The mains connections shown in Fig. 1 must be strictly adhered to. In Fig. 2, mains line should be the connection to FS1; neutral to S1b. Reversal of these connections could result in live mains potential appearing at SK2.

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TRANSISTORS AC127 Tp BFX29 Bp AC176 Tp BFX84 Bp AC176 Tp BFX84 Bp AC176 Tp BFX86 Sp AC178 Tp BFX86 Sp AC178 Tp BFX88 Sp AC186 Tp BFX81 Sp AC187 Tp BFY51 Tp AC187 Tp MAT100 Sp AD161/162 Tp MAT101 Sp AD161/162 Tp OC23 4p AF125 19p OC35 4p AF126 20p OC23 4p AF126 10p OC245 12p AF178 67p OC21 12p AF179 66p OC72 12p AF178 67p OC200 Tp BC107 12p ST141 3p BC108 11p OC201 3p BC117 12p Nt696 15p BC1617	ELECTROLYTICS μf 4500 19p 1.000 μf 25V 37p μf 3500 14p 2.000 μf 50V 36p $8\mu f$ 350V 14p 2.000 μf 50V 35p $8\mu f$ 450V 16p 2.000 μf 50V 35p $16\mu f$ 450V 16p 2.000 μf 55V 35p $25\mu f$ 50V 8p 3.000 μf 25V 35p $50\mu f$ 50V 8p 3.000 μf 25V 35p $50\mu f$ 50V 10p 8- μf 450V 27p $250\mu f$ 50V 10p 8- μf 450V 27p $250\mu f$ 50V 12p 16- $16\mu f$ 450V 27p $250\mu f$ 50V 12p 16- $16\mu f$ 450V 27p $250\mu f$ 50V 12p 16- $16\mu f$ 470V 7p $250\mu f$ 15V 7p 350\mu f 15V<	PLUGS Car aerial Co-axial D.I.N. 2 pin (speaker) D.I.N. 5 pin, 180 D.I.N. 5 pin, 180 D.I.N. 5 pin, 240 D.I.N. 5 pin, 240 D.I.N. 5 pin, 240 D.I.N. 5 pin, 180 d.AK. 34mm unscreened d.AK. 34mm unscreened d.AK. 34mm unscreened
 20Ω volume control for 3Ω speakers, 20p Antex CN240, ISW miniature soldering iron, 41-70 Valve and Transistor Data book, 9th edition, 75p Transistor equivalent book, BPI, 40p Transistor equivalent book, BPI, 40p Panel fuseholders 12 inch, 18p Panel fuseholders 20 Antex, 15p Packet of 24, 6 B.A. nuts, ½ in. bolts and washers, 16p Single, less switch, 15p Single, D.P. switch, 24p Tandem, less switch, 40p SkΩ, 10KΩ, 25kΩ, 50kΩ, 100kΩ, 250kΩ, 500kΩ, IMΩ, 2MΩ SLIDER CONTROLS, 87mm. complete with knobs, Single, 44p; Tandem, 55p, 10kΩ, 25kΩ, 50kΩ, 100kΩ, log, or lin. RESISTORS Carbon All 5%, high-stability, E12 values. W. 1p; 10W, 12p LOW-OHM RESISTORS Q:40, 35Ω, 30Ω, 47Ω, 56Ω, 68Ω, 82Ω, 10p each. TYGAN top quality loudspeaker covering material. Please send 4p for samples, sizes and prices. 	VEROBOARD Size 0-1 matrix 0-15 matrix 24:0: 34:0 12p 14p 24:0: 34:0 12p 14p 24:0: 34:0 12p 14p 24:0: 34:0 12p 14p 24:0: 34:0 12p 14p 25:0: 34:0 12p 14p 27:0: 44:0 12p 14p 14p 14p 14p 14p 14p 14p 14p 14p 14	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$



NEWS BRIEFS

Into the Light

A NOTHER company announces its move into that "most exciting field for the next decade", opticalelectronics. FR Electronics of Wimborne, Dorset, was formed in 1963 as a division of Flight Refuelling. First specialising in the manufacture of reed switches, FR Electronics has since expanded its interest in timers, special relays, and a.c. controllers. Now the Company is moving into this newest emerging market with a broad range of opto devices, including visible light sources and infra-red emitters, solid state detectors, optical coupled isolators, and digital readout displays.

In the main, FR Electronics will be marketing devices manufactured by other firms, but they have entered into an agreement with Opto-Electronics of Dallas, Texas. U.S.A. to manufacture this American Company's products under licence.

Computer-Aided Design

THE second conference and exhibition on computer aided design CADEX '72 was held at Southampton University during April. The joint organisers were, as in 1969. The Institution of Electrical Engineers and the Electronics Engineering Association.

Foremost among the exhibitors were Redac Software Ltd, who specialise in design services for the electronics industry. Computer Programmes (Software) are provided for use in customers' computers, bringing a powerful design tool to the aid of any establishment with suitable computer facilities. The programmes cover: *logic simulation*, so that an inter-connected system can be adjusted and modified until satisfactory operation is achieved. Circuit Analysis, of large networks. Circuit Layout for Printed Circuit Board Design and for microelectronics. Key feature of this software is its use of interactive graphics to ensure that the optimum results are achieved by the interaction of the designer with the computer via a light pen.

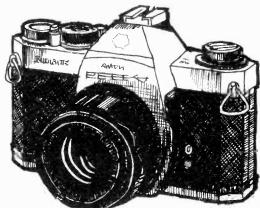
Computer-Aided Design (CAD) is not only a tool for the electronics industry. This was emphasised by the exhibit of DA Computer Services (part of an international engineering organisation involved in the design and construction of industrial installations for manufacturing industries). This firm provides a similar service for mechanical and structural engineers. The design of gear wheels, the layout of pipe systems, calculation of stresses in beams and frames, are some typical examples of the engineering programmes supplied.

Automated draughting systems are an essential part of CAD. Flatbed plotters can draw complex wiring diagrams or P.C.B. artwork masters in minutes, from data stored in a computer. Calcomp Ltd. demonstrated their latest model, the 7800 flatbed system which can ink write on paper or mylar, and scribe or cut on two layer films.

The Ferranti Interactive Freedcraft System produces input data for controlling automatic draughting machines or numerically controlled machine tools. It uses a "digitiser" to convert graphical or pictorial information into digital form for computer input. This operates on an electro-magnetic system of position measurement, offers no drag to manual movements, and gives the operator complete freedom to follow complex curves or irregular shapes smoothly and accurately.

Many CAD users require access to remotely located computers, and so it was no surprise to see the Post Office in attendance. Circuits for computing data transmission are provided by the Post Office and a number of different services are offered, via the public telephone network, the Telex System, or by telegraph private circuits.





Camera Shutter Tester

In many cases, exposure errors can be accredited to camera shutter faults, even in expensive professional cameras. Variations of time, as much as 50 per cent, can be measured from 1 millisecond (1/1,000) to 10 seconds.

Oscilloscope Pre-amplifier

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also Part 2 of the P.E. Digi-Cal commences constructional details.





August issue on sale July 14



Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

Fit for the mind

Sir,—With reference to Mr D. Bollen's letter on "Brainwave Reinforcement" I find that I must strongly disagree with the points he raised.

Firstly, he declares that Mr Brown's idea does not work. Well, I built the circuit and (although not exactly as described in the article) obtained a reasonable degree of success.

Secondly, Mr Bollen says alpha waves are produced by all normal persons. In fact, approximately 20 per cent of the "normal" persons do not produce alpha waves at all.

And finally he says that the subject must have his eyes closed and his mind blank to produce these waves. If Mr Bollen had watched the BBC programme *Horizon* he would have seen a young boy being taught how to prevent epileptic fits by CONCENTRATING on a row of lights and lighting successive bulbs by alpha wave production.

I hope all the other readers of Mr Bollen's letter have not been discouraged from building Mr Brown's very interesting idea. K. Reed.

Northumberland.

Matter of opinion

Sir,—On reading my May issue of P.E. I was very interested in the book review of "Guide to Printed Circuits" by G. J. King.

I am in complete disagreement with your reviewer's findings.

I have read every one of Mr King's books both published under his name and his Nom-de-plume and I thought that this book was his best practical book by far. The information on printed circuit substitutes was alone worth the cost of the book. The description of soldering and de-soldering methods was practical and informative.

Your reviewer forgot that a soldering iron is the basic tool of the enthusiastic amateur and the better the tools we have the better the finished job contrary to the opinions of Mr Lewis, (was it not) I consider that this is a *must* for the serious constructor and is a worth while contribution to the literature of the subject.

I have no connection with Mr King and I have never met or seen him.

> D Fisher, Middlesex

Required by law

Sir,—Having been a caravanner for some 20 years, I was particularly horrified to read Mr L. Musworth's suggestion in the *Ingenuity Unlimited* pages of your April edition of P.E.

The heavy duty flasher units that one buys for use on caravans and trailers have an extra terminal, which is connected to an extra warning lamp, which, when mounted within view of the towing vehicle's driver, flashes *only* when the load of a third (trailer flashing indicator) lamp is connected in parallel to the lamps of the towing vehicle.

This warning lamp or an audible system is required by law, and must indicate that the filament of the trailer's flasher is not open-circuit. It is not enough to know that volts are being sent to it, by connecting an indicator lamp in parallel with the trailer flasher filament.

The other drawback with the circuit is that the vehicle's flasher un't is used in series with the front and rear flasher lights, which are returned to earth at the respective sockets. This then would necessitate connecting an extra "live" lead from the fuse box (at the front of most vehicles), to an auxiliary set of relay contacts.

This circuit is, however. superfluous, as it still breaks the law. I am afraid that all this means that one must buy the correct item whatever the price!!!

W. A. Rawcliffe, Teddington.

Carbon leads

Sir,—With reference to the letter from Mr H. D. Briggs, in the May issue of PRACTICAL ELECTRONICS, I am well aware that the majority of modern cars are fitted with carbon string h.t. leads. My own car is fitted with such leads and has run on the "Scorpio" system for over 12,000 miles without any trouble whatsoever. Trouble will only occur if the leads are in poor condition, and if this is the case they should be replaced whatever system used.

I sympathise with his concern for the reliability of car electronics, but he isn't likely to get any useful opinion from a car electrician. Most of these have no knowledge of electronics beyond that required to change a bulb.

D. S. Gibbs.

NOTE: We regret that due to a printer's omission, the formula at the end of Mr G. A. Cozens' letter last month was not complete. The last line should read:

$$R = \frac{R3}{R4} \times R2$$

BACK NUMBERS WANTED Anyone who can supply the undermentioned are asked to communicate directly with the reader. We regret that back numbers of Practical Electronics can no longer be supplied. We will try to publish announcements of readers' requirements (without a guaranteed date) free of charge.

January, February, March, April and July, 1971 Mr. A. M. Cash, 5, Codrington Road,

Mr. A. M. Cash, 5, Codrington Road, Bristol, BS7 8ET.

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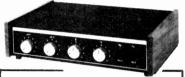
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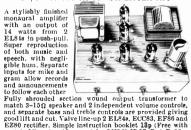
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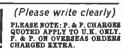
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Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

Project 60 modules are more versatile – using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all – price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

Typical Project 60 applications

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Project 60 Stereo F.M. Tuner



£9.98

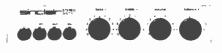
£5.98



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an LC. In the specially designed stereo decoder and squelch circuit for silent tuning between stations. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity: 7/4V for lock-in over full deviation. Squelch level: 20/4V. Signal to noise ratio: >65dB. Audio frequency response: 10 Hz – 15 KHz (±1dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: 2/4V. Cross talk: 40dB. Output voltage: 2 x 150mV R.M.S. Decention voltage: 2 x 150mV R.M.S. Operating voltage: 25-30 VDC. Indicators: Stereo on ; tuning. Size: 93 x 40 x 207mm.

Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

SPECIFICATIONS—Input sensitivities: Radio – up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A curve ±1dB:20 to 25.000 Hz. Ceramic p.u. – up to 3mV: Aux – up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE + 12 to -12dB at100 Hz. For the table of table of

A.F.U. High & Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current – 3mA H.F. cut-off (--3dB) variable from 28KHz to 5KHz. L.F. cut-off (--3dB) variable from 25Hz to 26KHz to 26K 100Hz. Distortion at 1KHz (35V. supply) 0.02% at rated output. Size: 66 x 40 x 90 mm.

Z.30 & Z.50 power amplifiers

a 6	5 6
and a second and a	a da ana a Reserva

Built, tested and guaranteed with circuits and instructions manual. 2.30 £4.48 z.50 £5.48

The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at 15w (80) and all lower outputs. Whether you SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications).

Power Outputs Z.30 15 watts R.M.S. into 8 ohms using 35 volts 20 watts R.M.S. into 3 ohms using 30 volts. **Z.50** 40 watts R.M.S. into 3 ohms using 40 volts : 30 watts R.M.S. into 8 ohms using 50 volts.

Frequency response: 30 to 300,000Hz±1dB.

Power Supply Units



Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ 6 where a stabilised supply is essential.

use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

Built, tested and guaranteed.

Built tested and guaranteed.

Oistortion: 0.02% into 8 ohms. Signal to noise ratio: better than 70dB unweighted. Input sensitivity: 250mV into 100 Kohms (for 15w into 8Ω) For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57 mm.

> PZ.5 30 volts unstabilised £4.98 PZ.635 volts stabilised F7 98 PZ.8 45 volts stabilised (less mains transformer) F7 98 PZ.8 mains transformer £5.98

Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail, Air-mail charged at cost.



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906A 2710 2N5355 2710 BC108 907 80012N5356 3210 BC109	10p BFY11 421p MPF102	4210 OC207 750 8710 OCP71 4210	VEROBOARD 0.15 0.1 Matrix Matrix	1·5μF 21p 2·2μF 25p
923 15p 2N5365 474p BC113 924 15p 2N5366 394p BC115	15p BFY18 32 p MPF104 15p BFY19 32 n MPF105	371p ORP12 50p 871p ORP61 50p	24in × 33in 17p 23p 23in × 5in 25p 25p	WIRE-WOUND RESISTORS
925 15p 2N5367 571p BC116A 926 2N5457 371p BC118	10p BFY20 #1.60 MPS3638	3210 P346A 2210	31in × 31in 25p 25p 31in × 5in 30p 29p	2.5 watt 5% (up to 270 ohn
ellow 121p 28005 75p BC121 ellow 121p 28020 £2 BC122	20p BFY24 45p NKT124 20p BFY25 25p NKT125	4210 TIP30A 60p 2710 TIP31A 6910	5in × 17in (plain) 83p	only), 7p, 5 watt 5% (up to 8.2k Ω only), 9 10 watt 5% (up to 25k Ω only
range 124p 28102 50p BC125 011 30p 28103 25p BC126	20p BFY26 20p NKT126	Z/19 T1P32A 75p	Vero cutter, 45p ; Pin insertion Tools (0.1 and 0.15 matrix) at 55p .	10 watt 5-8 (np to 25k ff only 10p.
014 324p 28104 25p BC140 053 18p 28501 324p BC147	371p BFY30 50p NKT135 10p BFY41 50p NKT137	2710 TIP34A #2-05	THE REPORT OF TH	POTENTIOMETERS
055 62p 28503 271p BC148		30p TIS43 40p	OPTOELECTRONICS MINITRON 3015F 7-SEGMENT	Log. or Lin., less switch, 16p. Log. or Lin., with switch, 25p.
133 30p 3N126 70p BC152	12p BFY50 23p NKT210 12p BFY51 20p NKT211 174p BFY52 23p NKT212 20p BFY53 174p NKT213	30p TIS45 27p	INDICATOR (SEE P.E. MARCH 1972) #2	Wire-wound Pots (3W), 38p. Twin Ganged Stereo Pots, Lo
136 25p 3N141 721p BC158 BC159	11p BFY53 174p NKT214 BFY56A 571p NKT215	2210 TIS46 11p 2210 TIS47 11p 2210 TIS48 121p	GNP-7AH COLD CATHODE SIDE-VIEW NIXIE TUBES 75p. TIL 209 RED LIGHT EMITT-	and Lin., 40p.
391 20p 3N143 671p BC160	891n DUVYO 30p NKT216	2239 11848 1239 3719 11849 1239 4239 11849 1239	TIL 209 RED LIGHT EMITT- ING DIODE (TEXAS INST.) 35p	PRESETS (CARBON)
391A = 30p = 3N152 = 874p = BC167 = 392 = 174p = R.C.A.; = BC168B = BC168B = 300 = 1000 = 1000 = 1000 = 1000 = 100000 = 100000 = 100000000		300	RESISTORS	0-1 Watt 6p 0-2 Watt 6p
393 15p 40050 55p BC168C 394 15p 40244 22 p BC169B	11p BPX25 \$1.85 NKT223 11p BPX29 \$1.80 NKT224	2710 T1852 1210 250 T1853 2210	Carbon Film	0-3 Watt 7fp
402 221p 40251 871p BC169C 403 221p 40309 321p BC170	12p BPY10 21.45 NKT225 124c BRY39 474c NKT229	221p T1860 221p 30p T1861 25p	1 watt 5%, 1p. 1W & 1W are 1 watt 5%, 1p. E24 series.	THERMISTORS R53 (STC) #1-20
404 821p 40310 45p BC171 Matching charge (audio	15p BSW41 423p NKT237 transistors only), 15p extra per pair.	35p T1862 271p	watt 5%, 11p. 1W, 1W, 4.2W 1 w M/O 2°, 4p. are E12 series. 1 watt 10%, 21p.	K151 (Ik) 12p VA1077 20 VA3705 95p
	ommonwealth (Air) Letter 65p (Min.) Alteration without prior notice.	. Farcel £1-69 (Min.),	1 watt 10%, 2 ¹ p. 2 watt 10%, 6p.	Mullard Thermistors also in stock Please inquire.
ALLERS WELCOME A. M	ARSHALL & S	ON Send Lists	15p for Comprehensive Price	MON-FRI 9-5.30
. 01-452 0161/2/3 A. T Telex 21492 28 CR	ICKLEWOOD BR	OADWAY.	LONDON, N.W.	
Telex 21492 25 CH	ICALEWOOD BK	CADWAT,	LUNDUN, N.W.	L 3AI 3-3



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220/240V. 50 cycle solenoid with laminated core so very silent in operation. Closes 4 with faminated core so very silent in operation. Closes 4 circuits each rated at 10 amps. Extremely well made by a German Electrical Company. Overall size 21 × 2 × 2iu. #1 oneb

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Double Leaf Contact. Very slight pressure closes both contacts. By each, 60p. doz. Plastic push-rod suitable for operating, 5p each, 45p doz.

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CAR AERIAL with dashboard control switch—fully extendable to 40in, or fully retractable. Suitable for 12V positive or negative earth. Supplied complete with fitting instructions and ready wired dashboard switch. **25**:75 plus 25p post and ins. $(\bigcirc$

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15 AMP FLECTRICAL .

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PROGRAMMER Learn in your sleep. Have radio playing and kettle boiling as you awake, switch on lights to ward off

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intruders, have warm house to come home to. All these and many warm house to come home to. All these and many other things you can do if you invest in an electri-cal programmer clock by famous maker with 15 amp on/off switch. Switch on time can be set anywhere to stay on up to 6 hours. Independent 60 minute memory jogzer. A beautiful unit. Price **\$1.95** + 200 μ , λ μ , or with glass front chrome bezel 75 μ , extra.



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Complete Kit (except wooden battens) to make the metal detector similar to the circuit in Practical Wireless August issue. \$2:95 plus 20p post and insurance.

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GUICK CUPPA Mini Immersion Heater, 350W, 200/240V. Bolis full cup in about two minutes. Use any socket or lamp holder. Have at beclside for tea, bay's food, etc. \$1.25, post and insurance 14p. 12V car model also available same price. Jugy heater \$1:50 plus p. & p. 14p.

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-STANDARD WAFER SWITCHES

CAPACITOR DISCHARGE CAR IGNITION

0	Standard	size l	} waf	er—silver	-plated	5-amp	contact
907. 9	standard	f" spind	lle 2″ 1c	ongwith	locking	washer	and nut
						1.0	. 1.1

No. of Poles	2 wav	3 way	4 way	5 way	6 way	S way	9 way	10 way	12 way
1 pole	40p	40p							
2 poles	40p	70p	70p						
3 poles	40p	40p	40p	40p	70p	70p	70p	95p	95p
4 poles	40p	40p	40p	70p	70p	70p	70p	£1-20	£1-20
5 poles	40p	40p	70p	70p	95p	95p	95p	£1.45	£1-45
6 poles	40p	70p	70p	70p	95p	95p	95p	£1.70	£1.70
7 poles	70p	70p	70p	95p	£1-20	£1.20	£1-20	£1-95	£1.95
8 poles	70p	70p	70p	95p	£1-20	£1.20	£1·20	£2-20	£2-20
9 poles	70p	70p	95p	95p	£1-45	£1.45	£1-45	£2-45	£2-45
10 poles	70p	70p	95p	£1.20	£1-45	£1-45	£1-45	£2.70	£2-70
11 poles	70p	95p	95p	£1-20	£1.70	£1.70	£1.70	£2-85	£2-95
12 poles	70p	95p	95p	£1-20	£1.70	£1.70	£1.70	£3-20	£3-20

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sitive or negative systems. V vehicles. **25-25** plus 20p.

ding probe tube and crystal earpiece. **£2**—twin stetho-set instead of earpiece **75p** extra—post and ins. 20p.

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S MONTH'S SNIF I3 AMP TWIN GANG SOCKETS Offered at less than wholesale price your opportunity to replace those dangerous adaptors-brown hakelite funch intonnting standard fitting. Unswitched 20p each, separately switched 30p each. Reparately witched and with neon on/off indicators 45p each. Less 10% ten or more + 20p postage if order under £5. order under £5.

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For any lamp up to 4kW. Mounted on switch plate to fit in place of standard switch. Virtually no radio interference Price **22:50**, plus 20p post and insurance.



MULLARD AUDIO AMPLIFIER MODULE

Uses 4 transistors, and has an output of 500mW into obms speakers. Input suitable for crystal mic, or pick-up, 9V battery operated. Size 2in long x 11 in wide x 1 in high, SPECIAL SNIP PRICE 60p each, 10 for 25.



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FURST FIANN "TIME & SET" SWITCH (A 30 Amp Switch.) Just the thing if you want to come house to a warm house without it costing you a fortnne. You can delay the switch on time of your electric fires, etc., up to 14 hours from setting time or you can use the switch to give a boost on period of up to 3 hours. Equally suitable to control processing. Regular price probably around £5. Special snip price £1-50. Post and ins. 23p.

MOTORISED COMPRESSORS

Ex mains operated frig. mils. There should be plenty of applications for these and we invite customers' suggestions. Our price is **£3 each**—carriage, etc., 50p. These are complete with starting relay and are sold as being in good order but untested.

HONEYWELL PROGRAMMER

HONEYWELL PROCRAMMER This is a dram type timing device, the dram being allibrated in equal division which are infinitely admistable for pa-terior and the second second second second second second type operated by the trips thus 15 circuits may be changed per revolution. Drive moto is maine operated 5 revs per min. Some of the many mass of this timer are Machinery control, Boiler fining. Hispensing and Vending machines, Display lighting an imated and signs, Signalling, etc. Price from makers probably over £10 each. Special snip price **25-75** pins 25p post and insurance. Don't miss this terrific bargain.

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INTEGRATED CIRCUIT BARGAIN A parcel of integrated circuits made by the famous Plessey Company. A once-in-a-lifetime offer of Micro electronic devices well below cost of manufacture. The parcel contains 5 ICs all new and perfect, first-grade device, definitely not sub-standard or seconds. 4 of the ICs are single silicon chip 6P ampilters. The 5th is a monolithic NPN matched pair. Regular price of parcel well over 55. Full circuit details of the ICs are included and in addition you will receive a list of many different IC's available at bargain prices 25p upwards with circuits and technical data of each. Complete parcel only 51 post paid. DON'T MISS THIS TERRIFIC BARGAIN.

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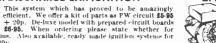


BATTERY CONDITION TESTER Made by Mallory but suitable for all batteries made by Ever Ready and others, most of which are zinc carbon types but also mercury manganese—nicad—silver oxide and alkaline batteries may be tested. The tester puts a dummy load on the battery and the meter scale indicates the condition depending upon which section the pointer rests. The section reads "replace" "weak" or "good". The tester is complete in its case, size 3jin \times 6 (in \times 2) in with leads and prods. Price 21.75 plus 20p postage.

Where postage is not stated then orders over £5 are post free. Below £5 add 20p. Semi-conductors add 5p post. Over £1 post free. S.A.E. with enquiries please.



ZPM-MODULATION MOTOR



Could also be used

Could also be used to open ventilatore, doors, valve, damp-er, etc., particulary suitable for remote control. Made by Satchvell. Easentially a rever-bile gazen motor fitted with internal init witches to stop it at the end of its travel. Size approx. 6in × 6 in × 5 in and weighing approx. 10b. This is extremely powerful and would lift a heavy door or open a long line of ventilators. To operate this motor you put the 50 cycle supply through a change over switch. For instance a thernostat with changeover contacts could automatically regulate the tem-perature in a growing house, chicken hatchery store star available revisitor, wirces from this to a volt meter would give a remote indication on the originous the state of open or close. Also internally witted is a variable revisitor, wirces from this to a volt meter would give a remote indication on the originous down Transformer is 21s. ChANGE-OVER THERMOSTAT

CHANGE-OVER THERMOSTAT

Satchwell type QX. This is ideal for working with the Satchwell ZPM. Extremely Satchwell ZPM. Ferrencip accurate in operation. Neat modern design in gold fnished perature setter. These are set for the range 10^{-257} C but the setting is adjustable by an internal screw to make it suitable for other temperature ranges. This is an expensive Thermo-stat. Our price however is only **\$4-50**.

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Made hy Smiths, these arc ac mains operated, NOT CLOCKWORK. Ideal for mounting on rack or shelf or can be built into box with 13A socket. 2 com-pletely adjustable time

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KIT FREE Eagle tional Japanese made these are ex-cellent value for not expect to be able to re-

able to repeat this offer once stocks are sold. Brief description of each kit is given below and with 3 kits or more we give FREE an accurate 11 piece balance kit Price of kits 40p each post paid. Special price for all 7 kits 82-50 with free balance kit. KA2 Lens kit. Eleven parts, including candle, one concave lens, one convex tens, stage and slit fraunc, etc. Watch light rays bend as they pass through different lenses.

frame, etc. Watch light rays hend as they pass through different lense. KA3 Water Pump Kit. Thirteen parts. Top of pump is transparent so that operating parts may be observed. Simall parts are brightly coloured to be seen easily while working. There types of pump may be made Lift pump, Farce Pump and Force Pump with reservoir and nozzle. KA4 Buzzer Kit. Eleven parts. Transparent covers allow the operation of buzzer to be seen. Illustrates and teaches how electromagnetism with an automatic switch results in an operating buzzer.

buzzer. KA6-3-Pole Motor Kit. Twenty-four parts, including enamel wire, armature and pole piece, etc. Motor operates from 11V battery. Illus-trates and teaches how electro-magnetism operates

KA7 Electro-Magnet Kit. Fifteen parts, includes KA7 Electro-Magnet Kit. Fifteen parts, includes compass. Makes two electro-magnets, one with one layer of wire and one with several layers of wire. Picks up tacks, nails and any small metal parts showing how magnetism works. KA8 Current and Registance Kit. Twenty-mine parts, including bench and light bulb. Conduct interesting and educational projects to learn the application of "OHMS LAW" and see the differ-

ence in current and resistance with different types

and lengths of wire. **KA9 Bell Kit**. Eight parts, including bell and push button switch. Build a complete electric bell and see how the hammer is triggered to make the bell ring.

PULSE GENERATORS

Sectronic made by Smiths. Operated by single 1-5V battery or transformer and rectifier. Two models, one gives 10 pulses per second the other gives 8. In plastic enclosures, size approx. 2in x 12in x 11 in deep. Price £2 each 10 for £18. MAINS TRANSFORMER SNIPS Wint Tenderer

Mains Transformer, Primary 240V tapped 220V. Secondary 20V (amp. Price 60p each or 10 for £5-40.

2040. Transformer. Primary 230-240V. Secondary 65-0-65 1 amp. With fitted primary «creen 65p each or 10 for 25-85.





PRACTICAL ELECTRONICS "SCORPIO" **ELECTRONIC IGNITION** SYSTEM



This Capacitor-Discharge Electronic Ignition system was described in the November and December issues of Practical Electronics. It is suitable for incorporating in any 12V ignition system in cars, boats, go-karts, etc., of either phe. or neg. earth and up to six cylind and components are required. Helps to promote easier starting (even under sub-zero conditions), improved acceleration, better high-speed per-formance, quicker engine warm-up and improved fuel economy. Eliminates exceessive contact-breaker point burning and the need to adjust point and spark-plug gaps with precision. Construction of the unit can easily be completed in an evening and installation studid take no longer than half an hour. A complete dwith each kit together with ready-drilled roller-tinned professional fusch-diele construction-ponents are available separately. Case ize 7jin x 4jin x 2 in approx. Complete assembly and wiring manual 25p, reflundable on purchase of kit. Price: £10:50 plus 50p P. & P. S.A.E. with all enquiries.

PSYCHODELIC LIGHTING UNIT Mk. 3

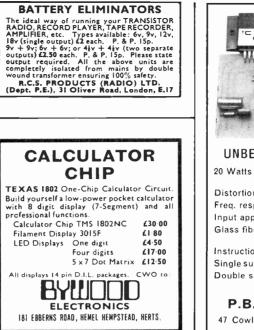


This unit represents a natural progression from our phenomenally successful Mk. I and 2 Units. As before the drive voltage is derived directly from the amplifier output or across the speakers. The unit converts the audio frequency signals into a threeing on the frequency of the signal and the intensity on the loudness of the audio source

he unit is constructed on professional fibre-glass printed-circuit board material and uses latest full-wave triac circuitry. There is a master-level control, together with independent sensitivity controls for each channel. The original minimum ambient light level controls have been redesigned permitting their use as faders; redesigned permitting their use as laders; allowing dimming from max, to zero at the turn of a knob. R.F.I. suppression is now incorporated as standard as well as provision for D.J. "Pulse-Flash" controls. The choice of two inputs enables operation from both high and low power amplifiers. Max, power I-SkW per channel at 240V a.c.

Complete assembly built and tested. Size $9in \times 7in \times 3in$. Price £25 carr. paid. S.A.E. with all induiries.

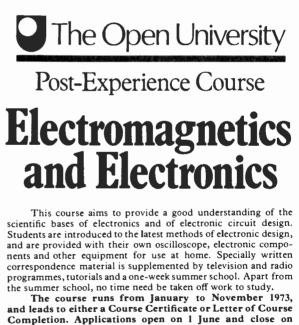
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NEW	STIC METE	W RS	'' DE	ESI	GNS NEL MET	! ERS
-A	TYPE SW. 100 x 80 m 500μA 1mA		TYPE 80 mm squar 50μΑ 50-0-50μΑ 100μΑ	re fronts £3-20 £3-10	mA	
$50\mu A \dots$ £3.6 $50 \cdot 0.50 \mu A \dots$ £3.4 $100 \mu A \dots$ £3.4 $100 \cdot 0.100 \mu A $ £3.3	50V d.e	£3·10 £3·10 £3·10 £3·10 £3·10 £3·10	100-0-100µ) 500µA 1mA 20V d.e 300V d.e	£8.00 £2.75	1A d.c 5A d.e 300V a.c	£2-60
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	41in 41in fronts. 50mA 1 100mA 1	2.80	Type MR	38P. 1 2	1/32in square 150mA	fonts. £1.60
50μΑ £3-60	500mA # 1A # 5A # 15A # 20V d.c. # 50V d.c. # 300V d.c. # 300V d.c. # 300V d.c. #	2-80 2-80 2-80 2-80 2-90 2-80 2-80 2-80 2-80			200mA 300mA 500mA 750mA 1A 2A 5A 10A	£1.60 £1.60 £1.60 £1.60 £1.60 £1.60 £1.60 £1.60
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	153 a.c. 2 300V a.c. 2 S Meter 1 ImA 2 VU Meter 2 IA a.c.* 2 5A a.c.* 2 10A a.c.* 4	2.87 3.60 2.80 2.80 2.80 2.80 2.80	50μA 50-0-50μA 100μA 100-100μA 200μA 500μA 500-0-500μA 1mA 1-0-1mA 2mA 5mA	. £1.90 £1.75 £1.75 £1.65 £1.60 £1.60	3V d.c. 10V d.c. 15V d.c. 20V d.c. 100V d.c. 300V d.c. 300V d.c. 500V d.c. 15V a.c. 50V a.c. 15V a.c.	#1 NO
Type MR.52P. 2 50μA £3·10 50-50μA £2:60 100μA £2:60 100-0-100μA £2:50	in square fronts. 20V d.c £ 50V d.c £ 300V d.c £ 15V a.c £	2.00 2.00 2.00 2.00 2.10	10mA 20mA 50mA 100mA	£1.60 £1.60 £1.60 £1.60	150V a.c. 300V a.c. 500V a.c. 500V a.c. 8 Meter 1mA VU meter	£1.70 £2.10
500μA £2:30 1mA £2:00 5mA £2:00 10mA £2:00 50mA £2:00 500mA £2:00 500mA £2:00 1A £2:00 500mA £2:00 1A £2:00 1A £2:00 1A £2:00 10V d.c. £2:00	1mA £	00.5	50μA 50-0-50μA 100μA 100-0-100μA 200μA 500μA 500μA 1mA 5mA	£2.25 £2.10 £1.87 £1.87 £1.75 £1.70 £1.70 £1.70	in square from 10V d.c 20V d.c 50V d.c 300V d.c. 15V a.c 300V a.c 8 Meter ImA VT Meter	£1.50 £1.50 £1.50 £1.80 £1.80 £1.80 £1.85 £2.25
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