## PRACTICAL



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Lets face it — an immediate success, the HY40 is here to stay HY40 means Hybrid Power, power neatly locked away inside an Intregrated Circuit. Power the modern way, simply mount only five additional components on a priated circuit board (all of which are supplied with the HY40). Power not only for $\mathrm{Hi}-\mathrm{Fi}_{\text {, }}$ power for Groups, for public address, for industry, power for all.
HY40 is HI-FI POWER ILP are POWER PROUD

In addition to the P.C. board and manual supplied with the HY40 we now include the five remaining components, at minimal cost, needed to complete the assembly of a High Performance Power Amplifier.
By merely combining two HY40s with a Stereo Preamplifier ( $2 \times \mathrm{HY}$ ) and simple Power Supply (PSU45), premium quality stereo may be obtained for a very modest outlay.
The free manual supplied with the HY40 gives clear, easy build instructions for Power Supply; volume, bass, treble and balance controls, together with inputs for Ceramic and Magnetic Pick-ups, Tape. Tuner and Auxiliary functions.
Internally the HY40 is based on conventional and proven circuit techniques developed over recent years.


OUTPUT POWER British Rating 40 WATTS PEAK, 20 watts RMS continuous.
LOAD IMPEDANCE 4-16 ohms INPUT IMPEDANCE 22 Kohms at 1 Khz .
INPUT SENSITIVITY 300 mV for maximum output.
VOLTAGE GAIN 30 db at 1 KHz . FREQUENCY RESPONSE 5 Hz $60 \mathrm{KHz} \pm 1 \mathrm{db}$.
TOTAL DISTORTION less than $1 \%$ (typical $0.1 \%$ ) at all output powers.
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The HY5 contains a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo preamplifier.

Specifically and critically designed to meet exacting $\mathrm{Hi}-\mathrm{Fi}$ standards, the HY 5 combines extremely low noise with a high overload capability. When used in conjunction with the HY40 and PSU45 forms a completely integrated system.

INPUTS
Magnetic Pick-up iwithin $\pm 1 \mathrm{db}$ RIAA curve) 2 mV .
Tape Replay (external components
to suit head). 4 mV
Microphone (flat) 10 mV
Ceramic Pick-up (equalized and compensatable) $\quad 20-2000 \mathrm{mV}$ variable.
Tuner (flat) 250 mV
Auxiliary 1250 mV
Auxiliary $22-20 \mathrm{mV}$
OUTPUTS
Main Pre-amp output 500 mV Direct tape output 120 mV .
ACTIVE TONE CONTROLS
Treble $\pm 12 \mathrm{db}$.
Bass $\mp 12 \mathrm{db}$.
INTERNAL STABILIZATION Enables the HY5 to share an unregulated supply with the Power Amplifier
SUPPLY VOLTAGE
15-25 volt
SUPPLY CURRENT
5m A approx.
OVERLOAD CAPABILITY
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input infinite on tuner and auxl.
OUTPUT NOISE VOLTAGE
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| $30 \mu \mathrm{~A}$ | 22.25 | 10 V d.c. | 11.50 |
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|  |  | SN74tiAN | $0.12$$0182$ | $\begin{gathered} 385 \\ 450 \end{gathered}$ | $\left\lvert\, \begin{aligned} & 25 \mathrm{Z4} \\ & 2525 \end{aligned}\right.$ | $\begin{aligned} & 30 \mathrm{p} \\ & 40 \mathrm{p} \end{aligned}$ | ELgjEMro |  |
| CA3n00 180p | FJH131 25p | 75p |  |  |  |  |  | $45 p$ |
| CA3005 117p | FJH141 25p | SN7442 75 | OZ4 | 30 p | 258 | 65 p | EM81 |  |
| CA3007 282p | FJH151 25p | 8N7446 100p | $1 \mathrm{L4}$ | 20 p | 30 Cl 5 | 80 p | EM84 | D |
| CA3011 750 | FJH161 700 | 8N7447 135p | 185 | 40 p | 31417 | 80 p | EM85 | 00 |
| CA3012 88p | FJH171 25p | 8N744＊125p | 185 | 30 p | 30 cis | 80 p | EM87 | 70p |
| CA3013 105p | FJH181 25p | SN7450 20p | IT4 | 25 p | 3975 | 850 | EY51 | 40p |
| CA3014 124p | FJH221 25p | － N 7451 | 114 | 30 p | 30 FLt | 75p | EY86 | 40 p |
| CA301\％84p | FJH231 25p | s： 7453 20p | IUs | 80 p | $30 \mathrm{FL1} 2$ | 120p | EY87 | 42p |
|  | FJH241 25p | 9N7454 20p | 2 D 21 | 35p | 3uFL14 | 95p | EZ40 | 55p |
| CA3018A ${ }_{110 \mathrm{p}}$ | FJH251 25p | 8N7460 20p | 3 Q 4 | 50 D | 30L15 | 85p | EZ 41 | 50 p |
| CA3019 84D | FJJ101 50p | 3N7472 30p | 384 | 35 p | 30 L 17 | ${ }^{80}{ }^{\text {p }}$ | EZ80 | 27p |
|  | FJJlll 80 p | 9，7473 40p | 3 V 4 | 48p | 30 P 12 | 80p | EZ81 | 29 p |
|  | FJJ121 60p | SN7474 40p | 5 R 4 | 75p | 30 P 19 | 85p | （iZ32 | 48p |
| $\begin{aligned} & \text { CA3020 } 126 \mathrm{p} \\ & \text { CA } 3020 \mathrm{~A} \\ & 180 \mathrm{p} \end{aligned}$ | FsJ131 60p | BN7475 45p | 5 O 4 | 35p | 30 PL | 75 p | OZ34 | p |
| CA3021 1560 | FJJ141 125p | 3N7476 45p | $5 \mathrm{~V}_{4}$ | 45p | 30 PL 13 | 93 p | K T6t | 05 |
|  | FJJ181 75p | SN7483 87p | 5 Y 3 | 40 p | 30 PLL 4 | 80 p | кт88 | 82.00 |
| $\begin{array}{ll} \text { CA3023 } & 126 \mathrm{p} \\ \text { CA302 } & 100 \mathrm{p} \end{array}$ | FJJ191 65p | SN7486 33p | 5241； | 40 p | $3{ }^{5} \mathrm{~L} 6$ | 50p | MU14 | 75 D |
|  | F．J3211 125p | SN7490 87p | 6／30L： | 80 p | 35 W 4 | 35p | PabC | 40p |
| $\begin{aligned} & \text { CA3026 } 100 \mathrm{p} \\ & \text { CA3028A } 74 \mathrm{D} \end{aligned}$ | FJJ251 125p | SN7492 87p | 6．act | 40 D | $3 \mathrm{ZZ4}$ | 35p | PCst | 60D |
| $\begin{aligned} & \text { CA3028A 74D } \\ & \text { CA3028B } \end{aligned}$ | FJLi01 125p | SN7493 87p | ${ }_{6 \text { fat }}$ | 40 p | 35 zs | 50 p | Pens | 0p |
| ${ }_{3029}{ }^{1057 p}$ | FJY101 25p | SN7495 87D | 6aks | 35p | эu¢ | 50 p | P¢97 | 45p |
|  | 1010 250p | sN7496 87p | 6akt | 60 p | 50c：， | 50 p | PC9\％ | 48p |
|  | $1 \mathrm{Cl} 2 \mathrm{250p}$ | SN74107 52d | galis | 20 p |  | 55 p | PCO 4 | 40p |
| CA3029．${ }_{165 p}$ | L900 40p |  | 6AM6 | 30 p | 45.1 | 50 p | $\mathrm{PCO}_{4}$ | 40 p |
| CA3030 137 p | L914 40p | 140p | 6.14 .5 | 38 p | 80 | 50 p | Pcuss | 650 |
|  | L923 | 8N74154 | $6^{6}$ ASt | 40 p | 162\％ | 50 p | PCC89 | 60p |
| CA 3036 | MC724P ${ }^{\text {60p }}$ | 220p | 8．AT6 | 35 p | 5783 | 70 p | PCCls9 | 85 |
| $\begin{array}{rr} \text { CA3039 } & 82 \mathrm{p} \\ \text { CA3041 } 109 \mathrm{p} \end{array}$ | MC780P 247p | SN74160 | bauti | 25p | ti4 | 180 | PCrso | 0 D |
|  | M ${ }^{\text {c788P }} 148 \mathrm{P}$ | －N74161 ${ }^{180 \mathrm{p}}$ | favi | ${ }^{30 p}$ | ${ }_{\text {A } 231}$ | 55 p | PCFs＇ | ${ }^{340}$ |
| $\begin{array}{ll} \text { CA3041 } & 109 \mathrm{p} \\ & \text { CA3042 } \\ \end{array}$ | MC799P ${ }^{\text {M }}$ M ${ }^{\text {M }}$ | 260 | 6BA6 6 BEF | ${ }_{30 \mathrm{p}}^{23 \mathrm{p}}$ | cras | 35 p 30 p | ${ }_{\text {PCFP8 }}$ | Op |
| CA3043 1370 | MC799 | SN74164 | ${ }_{6 B 146}$ | 75 p | Dat96 | 45 p | PCF800 | 80 |
|  | MC1303L | 20 | ${ }_{6}^{6 B J 6}$ | 50 p | $1 \mathrm{~F}^{91}$ | 22p | PCProl | 50 D |
|  | ${ }^{100} \mathrm{p}^{\text {1 }}$ | SN74165 | ${ }_{6 B 47}$ | 40 p | DF913 | 45 p | PCF802 | 80p |
| $\begin{aligned} & \text { CA3046 } \\ & \text { CA3047p } \\ & \text { 8187p } \end{aligned}$ | MC1304P | 22 | $6 \mathrm{BR7}$ | 90 D | Dк91 | 40 p | PCF805 | B |
|  |  | 8N7＋192 | 613 RY | 70 p | DK92 | 55p | PCFS06 | 寿 |
| $\begin{aligned} & \text { CA3048 } 204 \mathrm{p} \\ & \text { CA3049 160p } \end{aligned}$ | 5 | 175 | ${ }_{6}^{6 B W 6}$ | 850 | DK96 | 50 p | PCF808 | D |
| $\begin{aligned} & \text { CA3050 185p } \\ & \text { СА } 3051 \quad 184 \mathrm{p} \end{aligned}$ | M | 3NT4193 ${ }_{1750}$ | 6867 61325 | 80 p | DL92 | 35 p | PCL82 | p |
|  |  | TAA241 ${ }^{175}$ | ${ }^{613285}$ | 40 p | ${ }^{1} \mathrm{DL94}$ | 48 p | PCL83 | 45 |
| $\begin{array}{ll} \text { CA305\% } & 185 \mathrm{p} \\ \text { CA3053 } & 48 \mathrm{p} \end{array}$ | MCl 435 P | 162p | 604 $6000 \%$ | 33 p 185 p |  | 45 4 | ${ }^{\text {PCLC84 }}$ | ${ }_{40 \mathrm{D}}^{45 \mathrm{D}}$ |
| CA3054 109p | 345p | TAAB42 | 6 CL ¢ | 50 p | DY46 | 32 p | PCL86 | 46D |
|  | MC155\％${ }^{\text {a }}$ | 150p | 0 CW | ${ }^{658}$ | DY87 | 33 p | PrL20 | 855 |
| CA3059CA3064180p120p | 481p | TAA243 150p | 6 F 1 | 62p | Esbec | 100 p | PL36 | 55D |
|  | MC1709C： | TaA263 75p | ${ }^{6 F 6 G}$ | 35 p | E180F | 100p | PL81 | 60p |
| $\begin{aligned} & \text { FC1101 } 185 \mathrm{P} \\ & \text { FCH1111 } 105 \mathrm{p} \end{aligned}$ |  | TAA293 97p | ${ }_{6}^{6 \mathrm{~F}} 13$ | 45 p | EAbcxo | 35 p | PL＊： | 5p |
|  | Mr | TAA300 175p | 6 F 14 | 70. | EAF42 | 35p | PL43 | 45 |
|  |  | TAA310 ${ }_{\text {T }}$ | ${ }^{6615}$ | ${ }^{85} 5$ | Eb91 | 200 | ${ }^{\text {PL84 }}$ | 408 |
|  |  | TAA320 72p | $6_{619}$ | 50 p | EbCd 1 | 55p | PL500 | ${ }^{750}$ |
| FCH141 105p FCH151 1050 | PA：30 140p | TaA350 175p | $6 \mathrm{~F}^{2} 3$ | $85 p$ | еbcha | 30 p | PL504 | 80 D |
| FCH151 105p <br> FCH161 50p |  | TAA435 147p | ${ }_{6}^{6 \mathrm{H} 6}$ | 17 p | EBF80 | 40 p | PY32 | ${ }^{65}$ |
|  | Pras37 210 P | T AAspl 132p | 6.34 | 50 p | EyFs3 | 40 p | PY33 | 8p |
| ${ }_{\text {FCH }}$ FCH181 105p | PA．246 160p | TAA522 360p | 60． | 25p | Ebrby | 32 p | PY\％0 | 40 p |
|  | PA424 ${ }^{\text {Pas }}$ | TAA530 495p | ${ }^{6} \mathbf{3} 515$ | 30 D | Eblal | ${ }_{60} 0$ | PY81 | 80 D |
|  | PAP264 190p | TAA811 445p | ${ }_{6}^{6.56}$ | 20 p | ECss | 60 p | PYRZ | 85 p |
| FCH201 130p | PARE65 200D | TAB101 97p | ${ }^{\text {fiJ7 }}$ | 45 p | ECxR | 80 p | PY83 | ${ }^{380}$ |
|  | SN7400 200 | TAD100 150 p | 6 KBG | 40 p | Eccio | ${ }^{850}$ | PY88 | 400 |
|  | SN7401 | TAD110 150 p | 6L6GT | 45p | ECC84 | 80p | PY800 | 40 |
| FCH231 150p | SN740：2 20 p | \＄14035） 150 p | 6 LLD | 50 p | ECC85 | 40 p | PY80 | s0p |
| FC．FCJ101 180p1500 | 8N7403 |  | 647 | 40 p | Eccrs | 40 p | U | 0 p |
|  | $\begin{array}{ll}8 \times 7404 \\ 8 \times 7405 & 20 \mathrm{p} \\ 80 \mathrm{p}\end{array}$ |  | 68.7 6867 | 40 4 | Ectr80 | 85 p | U | 40 |
| F＇CJI31 2750 | SN740日 800 | UA703C 137p | ${ }_{6817}$ | 40p | ${ }_{\text {ECFP8 }}$ | ${ }_{65 p}^{35 D}$ | U50 | 40p |
| $\begin{aligned} & \text { FCJIA1 } \\ & \text { FCJ:001 } \end{aligned}$ | 8N7408 20p | UA709C 1259 | $6 \mathrm{KK7}$ | 40p | ${ }_{\text {ECH21 }}$ | 570 | ${ }_{\text {V191 }}$ | ${ }_{750}$ |
|  | 8N7409 20p | Ua710C 125 p | 68 L 7 | 35p | ECH35 | 100 p | ${ }^{1281}$ | 40 D |
|  | 8N7410 20 p | UA716 187 D | 6sN7 | 35 p | ECH42 | 750 | U 28.2 | 40 D |
|  | 8N7411 ${ }^{\text {S }}$ | UA723C 182 p | 6847 | 40 p | ECHA1 | 30 p | U301 | 0 p |
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|  |  |  | 6 vga | 25 p | ECL30 | 45 p | Uabeso |  |
| BRIDGE ${ }^{\text {Bras }}$ |  |  | 6 6 60 | ${ }^{32} \mathrm{D}$ | ECLb： | 35 p | UaF42 | 85p |
|  |  |  | $6 \times 4$ | 35 p | ECL43 | 70 p | UBC41 | 80 p |
| plastic | 200 PI | Y 4A 70p | ${ }_{6}^{6 \times 565}$ |  | ECL37 | ${ }^{40 \mathrm{P}}$ | UBCK1 UBFPo | 40 p |
| ENCAPSULATED600 PIVS0p | TED 400 PI | $\mathrm{V}^{4 \mathrm{~A}} 75 \mathrm{p}$ | ${ }^{6 \times 56}$ | 40 p | ${ }^{\text {EFP37A }}$ | 120 p 50 p | ［7BF89 | ${ }_{850}$ |
|  | 50 p ． 50 PI | －6A 62p | ${ }_{\text {lor }}^{10 \mathrm{c}} 1$ | 50p | Er＋4 | ${ }_{50 \mathrm{p}}^{50 \mathrm{p}}$ | U8F89 | 85 D 49 p |
| 50 PIV 2 A | 550 | V0．${ }^{0.45}$ |  |  | EF＋1 | 65 p | UCCs5 | 40 p 40 p |
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| $400 \cdot 2100 \mathrm{PIV}$ | 7 D | $\mathrm{pp}^{0} \quad 20 \mathrm{p}$ | 19AQs | ${ }^{35 p}$ | EH90 | 40 D | UF89 | 408 |
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| 4004400 PI | $8 \mathrm{p} \quad 10$ | D 25p | 20F＇S | 65p | ELL33 | ¢1．25 | Ules， | 40. |
| 40056001 1 1 | 10 p | 2p $\quad 26 \mathrm{p}$ | 20 LI | 21.10 | EL4 | ${ }^{60 p}$ | UY4t | 48 p |
| 4005800 PlW | 12 p | 50 27p | 20 P 1 | 50 p | EL4 | 85 p | UY83 | 400 |
| 400． $100+$ less $15 \% 100+$ less $20 \%$ |  |  | $20 \mathrm{P}^{3}$ | ${ }^{60 p}$ | ELAI | 55 p | VR105／3 |  |
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|  |  |  | 25L6 | ［）p | L91 | 35p | Sor pinta |  |
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|  |  |  | AAZ13 | 10 p | BY100 | 15p | 0.447 | 10 D |
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|  | ${ }^{2}$ | ${ }^{5}$ | ${ }^{4}$ | $\mathrm{EP}_{5}$ | t |
| 50 | 0.23 | 0.25 | 5 |  |  |
| 100 | 0.25 | 0.33 | 0.53 | 0.68 | 0.83 |
| 200 | 0.85 | 0.37 | 0.57 | 0.61 | 0.75 |
| 400 | 0.43 | 0.47 | 0.67 | 0.75 | 0.93 |
| 600 | 0.63 | 0.57 | 0.77 | 0.87 | 1.25 |
| 800 | 0.68 | 0.70 | 0.90 | 1.20 | 1.50 |

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 2p | \& p | \&p | 8 t | fp | \% | 8 p |
| 50 | 0.04 | 0.05 | 0.05 | 0.07 | 0.14 | 0.21 | $0 \cdot 47$ |
| 100 | 0.04 | 0.08 | 0.05 | 0.13 | 0-16 | 0.23 | 0.75 |
| 200 | 0.05 | 0.09 | 0.68 | 0.14 | 0.20 | 0.24 | 1.00 |
| 400 | 0.06 | 0.13 | 0.07 | 0.20 | 0.27 | 0.37 | 1.25 |
| 600 | 0.07 | 0.16 | 0.10 | 0.23 | 0.34 | 0.45 | 1.85 |
| 800 | $0 \cdot 10$ | 0.17 | 0.13 | 0.25 | 0.37 | 0.55 | 2.00 |
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| 1200 | - | 0.33 |  | 0.33 | 0.57 | 0.75 |  |
|  | TRI | IACS |  |  | AS | 10 |  |
| VBom | M 2 A | 6A | 10.4 |  | ECTI | IERS |  |
|  | TO-1 T | T0-66 TO | T0-88 | 35 a | 409 | PI.V' | Stud |
|  | Ep | \% | fp |  |  |  |  |
| 100 | 0.80 | 0.63 | 1.00 |  | DIA |  |  |
| 200 | 0.70 | 0.90 | 1.25 | FOR | Cs |  | ITH |
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| 1N85 | 0.88 | AC128 | 0.20 | BF184 0.20 | GJ7M | 0.87 | OC44 ${ }^{\text {a }}$ | 0.17 |
| 1N253 | 0.60 | $\mathrm{ACl}^{-}$ | 0.25 | BF185 0．80 | HG100J | 0.50 | OC4s | 0.12 |
| 1N206 | 0.50 | AC188 | 0.25 | BF19 0.17 | lis 100 A | 0.20 | OC45M | 0.18 |
| 1N645 | 0.25 | ACY 17 | 0.30 | BF195 0.15 | Mati00 | 0.25 | OC46 | 0.27 |
| 1N726A | 0.80 | ACY18 | 0.85 | BF196 0.15 | MAT101 | 0.30 | OC5 ${ }^{\circ}$ | 0.60 |
| 1N914 | 0.07 | ACY 19 | 0.25 | BF197 0．15 | MAT120 | 0．25 | OC5 | 0.80 |
| 1N4007 | 0.80 | ACY20 | $0 \cdot 20$ | BFS61 0.28 | MATl2l | 0.80 | 0 C 99 | 0.85 |
| 18021 | 0.80 | ACy21 | 0.20 | BFS96 0.28 | MJE540 | 0.87 | OC66 | 0.50 |
| 18113 | 0.15 | ACY22 | 0.10 | BFX12 0.20 | MJE 2950 | 1.87 | $0<70$ | 0.12 |
| 18130 | 0.18 | ACY27 | 0.25 | BFX13 0.25 | MJE3055 | 0.87 | Oc71 | 0.12 |
| 18131 | 0.18 | ACY28 | 0.17 | $\begin{array}{ll}\text { BFX } 29 & 0.25\end{array}$ | NK T128 | 0.85 | 0 OCF | 0.20 |
| 18202 | 0.28 | ACY 39 | 0.50 | BFX30 0．25 | NKT129 | 0.80 | 0 C 73 | 0.80 |
| 20240 | 1.97 | ACY40 | $0-15$ | BFX35 0.98 | NKT211 | 0.25 | 0 O 74 | 0.80 |
| $2 \mathrm{GS01}$ | 0.20 | ACY41 | 0.15 | BFX63 0.50 | NKT213 | 0.25 | $0<75$ | 0.85 |
| 2 Gs 02 | 0.82 | ACY44 | $0-25$ | BFX84 0.25 | NKT214 | 0.15 | $0 \mathrm{OC76}$ | 0.85 |
| 29306 | 0.80 | AD140 | 0.50 | BFX 85 | NKT216 | 0.37 | OC7 | 0.40 |
| $2 \mathrm{G371}$ | 0.22 | AD149 | ． 0.50 | BFX86 0.85 | NKT21\％ | 0.35 | OC78 | 0.20 |
| 29381 | 0.85 | AD161 | 0.87 | $\begin{array}{ll}\text { BFX } 87 & 0.25\end{array}$ | NKT218 | 1.18 | OC79 | 0.22 |
| 2 G 414 | 0.80 | AD16 | 0.37 | BFX88 0.20 | NKT219 | 0.38 | 0 C 81 | 0.20 |
| $2 \mathrm{G417}$ | 0.22 | AF106 | 0.30 | BFY10 1．00 | NKT224 | 0.20 | OC811） | 0.20 |
| 2N214 | 0.48 | AF114 | $0 \cdot 25$ | BFY11 1.25 | NKT224 | 0.82 | OC81M | 0.20 |
| 2N247 | 0.85 | AF115 | 0.25 | $\begin{array}{ll}\text { BFY17 } & 0.26\end{array}$ | NKT251 | 0.24 | OC81DM | 0.18 |
| 2N250 | 0.60 | AF116 | 0.25 | $\begin{array}{ll}\text { BYF18 } & 0.26\end{array}$ | NKT271 | 0.25 | $0 \mathrm{C817}$ | 0.40 |
| 2 N 404 | 0.20 | AF117 | 0.25 | BFY19 0.25 | NKT272 | 0.25 | OC82 | 0.25 |
| 2 N 697 | 0.15 | AF118 | 0.62 | BFY24 0.45 | NK T273 | 0.15 | OC821） | 0.20 |
| 2N698 | 0.40 | AF119 | 0.20 | BFY4 1.00 | NKTa゙ 4 | 0.20 | $00^{0} 83$ | 0－25 |
| 2N706 | 0.10 | AF124 | 0.25 | BFY50 0.28 | NKT275 | 0.25 | 0 O 84 | 0.25 |
| 2N706A | 0.12 | AF125 | 0.20 | $\begin{array}{ll}\text { BFYJ1 } & 0.20\end{array}$ | NKT9T7 | 0.80 | OC114 | 0.88 |
| 2N708 | 0.15 | AF126 | 0.17 | $\begin{array}{ll}\text { BFY52 } & 0.22\end{array}$ | NKT278 | 0.25 | OC122 | 0.60 |
| 2N709 | 0.88 | AF127 | 0.17 | BFY53 0.17 | NKT301 | 0.40 | OC123 | 0.65 |
| 2N711 | 0.87 | AF139 | 0.30 | BFY64 0.42 | NKT304 | 0.75 | OC139 | 0.25 |
| 2N987 | 0.58 | AF178 | 0.55 | BFY90 0.85 | NKT403 | 0.75 | OC140 | 0.85 |
| 2N 1090 | 0.80 | AF179 | 0.65 | Bsx2 0.50 | NKT404 | 0.55 | OC141 | 0.60 |
| 2N1091 | 0.88 | AF180 | 0.52 | BSX60 0.98 | NKT678 | 0.80 | OC169 | 0.20 |
| 2N1131 | 0.25 | AF181 | 0.42 | BSX76 0.15 | NKTil3 | 0.25 | Oc170 | 0.25 |
| 2N1132 | 0.25 | AF186 | 0.40 | BSY 260.18 | NKTi73 | 0.25 | OC171 | 0.80 |
| 2N1302 | 0.18 | AFY19 | 1.13 | $\begin{array}{ll}\text { B8Y07 } & 0.17\end{array}$ | NKTケT | 0.88 | OC200 | 0.40 |
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| 2N1308 | 0.25 | ASY27 | 0.32 | BT102／500R | OA47 | 0.10 | OC204 | 0.40 |
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| 2N2926 | 0.10 | BC149 | 0.15 | $\begin{array}{ll}\text { DD006 } & 0.18\end{array}$ | OAZ223 | 0.45 | XA151 | 0.15 |
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| 2N3702 | 0.10 | BC160 | 0.68 | $\begin{array}{ll}\text {（iD3 } & 0.33\end{array}$ | OAZ2442 | 0.28 | XA164＇ | 0.25 |
| 2 N 3705 | 0.10 | BC169 | 0.13 | $\begin{array}{ll}\text { Cil } \\ \text {（1）} & 0.05 \\ \end{array}$ | 9AZP44 | 0.82 0.29 | XA102 | 0.25 |
| 2N3706 2N3707 | 0.88 0.12 | BCY31 | 0.35 | $\begin{array}{ll}\text {（11）5 } & 0.33 \\ \text { G118 } & 0.25\end{array}$ | OAZ246 OAZ290 | 0.23 0.38 | X 1101 X13102 | 0.28 0.10 |
| 2N3707 | 0.12 0.10 | BCY3： | 0.55 | $\begin{array}{ll}\text { G108 } \\ \text {（1D1\％} & 0.25 \\ & 0.05\end{array}$ | ${ }^{0} \mathrm{Cl} 1690$ | 0.38 0.50 | XB10＇s | 0.10 |
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| 2 N 5027 | 0.58 | BCY4： | 0.85 | GET116 0.50 | OCu4 | 0.60 | ZS2T1 | 0.18 |
| 2 N 6088 | 0.88 | BCY ${ }^{\text {a }}$ | 0.15 | OET120 0．25 | OC2S | 0.37 | ZT以1 | 0.25 |
| 28005 | $1{ }^{1.00}$ | BCY71 | 0.80 | $\begin{array}{ll}\text { GET872 } & 0.30 \\ \text { GET873 } & 0.25\end{array}$ | OC26 | 0.25 | ZT43 | 0.25 |
| 28178 28301 | ${ }^{1}$ on ${ }^{\text {a }}$（ ${ }^{\text {a }}$ | BCZ10 BCZ11 | 0.35 0.50 | $\begin{array}{ll}\text { GET875 } & 0.25 \\ \text { GET880 } & 0.87\end{array}$ | OC24 | 0.80 | ZTX107 | 0.15 |
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|  | positive AND gate <br> Tosem pole ourput |  | 25p | $21 p$ | 18p | 7470 | J-K flip flop | FLJ 101 | 45p | 37 p | 32p | 74141 | vibrator <br> BCD to decimal | 101 | 48p | 40p | 34p |
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|  | gate | 121 | 20p | 16p | $14 p$ | 7476 | Dual J-K master. |  |  |  |  |  | Synchronous up |  |  |  |  |
| 7430 | 8-input NAND gate | 131 | 20p | $16 p$ | $14 p$ |  | slave flip-flop with |  |  |  |  |  | down 4-bit binary |  |  |  |  |
| 7440 | Dual 4-input NAND buffer | 141 | 24p | 20p | 17p |  | preset and clear | $\begin{aligned} & 131 \\ & \text { FLH } \end{aligned}$ | 45p | 40p | $36 p$ |  | counter with one line mode control | 211 | 61.80 | ¢1.48 | ¢1.27 |
| 7442 | BCD to decimal |  |  |  |  | 7480 | Gated full adder | 221 | 67p | 56p | 48p | 74192 | Synchronous up |  |  |  |  |
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## TRADE WINDS

FOR some time it has been clear that certain sections of the U.K. electronics industry will face a difficult. if not threatening, situation due to overseas competitors, chiefly Japan, in the coming years. For the home producers of television receivers, it seems the danger is quite imminent.

Shut out of the American market by recently imposed tariffs and debarred from Germany and Italy by the terms of the licensing agreement which allows them to manufacture television receivers for the PAL system. the Japanese look now to the large U.K. market as a main outlet for their huge output of colour receivers. And with 20 in models likely to retail at $£ 100$ against home products costing around $£ 180$. it is too much to expect patriotism to win against price. Anyhow precedents have already been provided by the watch and camera trades, and imported radio, audio, and measuring equipments are now taking over sections of the home electronics market.

There is, unfortunately, the possibility of an even more serious threat to the electronics industry in the next decade. The danger of an unimpeded flow of cheap products is not merely the worry of that part of the industry catering for the consumer market. Producers of industrial equipment may well have to face intense overseas competition in the coming years. Japan, for example, is reported to be making phenomenal progress in industrial electronics. She has already taken over the U.S.A. market for desk calculators; and as the largest ship building country in the world, she can be expected to exploit this situation fully, and to offer in time vessels fully equipped with communications, radar, and navigational equipment, as complete going concerns. This kind of development in world trade could be a serious blow to the capital equipment section of the U.K. electronics industry-the very section which has so far been the major exporter, in terms of value, of the whole U.K. electronics industry.

Can any measures be taken to avert the more serious consequences of this "cut price" competition from countries operating under quite different social and economic systems to our own? British industrialists no doubt look with envy to the U.S. and would welcome a similar initiative from our government. But tariffs and trade protection arouse no enthusiasm among customers at large. More likely our industry"s chief hope rests in the continuing expansion of world requirements for electronic products of all kinds. But frequent "priming of the pump" by the introduction of technical innovations that lead to further new and commercially valuable applications for electronic techniques seems now. more than ever, an imperative operation for the industry.

## THIS MONTH

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Our February issue will be published on
Friday, January 14.

[^1]
# Photo-Print Process By A.WOODROW 

MANY amateur photographers like to do their own processing, and an increasing number are attempting colour printing. One of the more tedious tasks of processing is the printing of a roll of negatives. When printing monochrome, each negative requires that a test strip be made to determine the correct exposure needed, the test then being assessed and a final print then made

When a roll of 36 negatives is being printed, the amount of time and wasted materials involved can be quite considerable. When colour negatives are used, the problem is multiplied by the necessity of obtaining the correct colour balance as well as density. It is possible to waste a complete printing session without obtaining a single satisfactory colour print.

The unit to be described is designed to take most of the trial-and-error out of printing: in fact, prints are made almost automatically in the case of monochrome. The work involved in colour printing is a little greater than for black and white, but it is still possible to make a successful print at the first attempt

## THREE FUNCTIONS

An exposure meter first assesses the negative, and the information obtained is transferred to a timer, which switches the enlarger lamp on for the required time to give a correctly exposed print. The exposure meter is also used for the comparison of a colour negative with a "standard" negative to obtain the correct colour balance.

One of the most important factors in successful photographic processing is consistency of conditions; temperatures, strength of solution, etc. must be constant if good results are to be obtained. One such factor which is often overlooked is mains voltage. If an enlarger lamp is operated at low mains.
its light will be more yellow than at nominal voltage. as well as giving a reduced light output. Conversely, at lamp operated at high voltage gives a bluer light.

The change in light output is serious enough in monochrome printing, but the additional variation in colour can cause chaos in colour printing; a change of two or three volts is sufficient to cause a noticeable shift in colour. With this in mind, the control unit includes a stabiliser unit to hold the lamp voltage constant against variations of mains

The control unit thus has three parts, intended to be used together, although any one or two sections could be constructed and used independently of the others. The unit will automatically assess and expose negatives over the range one second to one minute, and the stabiliser holds the lamp voltage constant for mains inputs between $220-250 \mathrm{~V}$. For nominal mains voltages other than 240 V , the stabiliser will operate over a correspondingly lower range.

## EXPOSURE METER

If a light source of known brightness is projected through a negative, the amount of light allowed to pass by the negative is a measure of the density of it. and thus of the exposure required to make a print. In this case, the light source is the enlarger lamp, and the amount of light passed by the negative is measured at the enlarger baseboard by a light dependent resistor

This device, as its name implies, varies its resistance according to the amount of light falling on its sensitive surface. When the device is placed on the baseboard, its resistance depends on the amount of light transmitted by the negative, and thus on the required exposure.

A light dependent resistor, however, has no "memory"; when the illumination is removed, the information gained from the measurement is lost.

## Control Unit

(3)



As this information is later needed by the timer. a further element must be added to store the reading. ready for transfer to the timer. A bridge is therefore used for the measurement, and the basic circuit is snown in Fig. 1.

## BRIDGE CIRCUIT

The main property of this circuit is that when $R_{1} \times R_{4}=R_{2} \times R_{3}$ no current will pass through the meter. In this condition. the bridge is balanced. When the equation is not satisfied, a current flows in the meter. This current is of a magnitude dependent on the degree of unbalance and the voltage of the supply, but the point at which batance occurs is independent of applied voltage.


Fig. 1. Basic bridge circuit

Referring now to the main circuit in Fig. 2, the bridge is formed by VR1, VR2 and the light dependent resistor PCCI. Resistors RIO, RII and R15 are also included in the bridge, but serve only to limit the range of the controls, preventing any arm of the bridge from being snort-circuited. Potentiometer VRI forms two arms, either side of its wiper, this control being preset according to the type of printing paper in use.

The supply is 12 V a.c., taken from the mains transformer. The output from the bridge is rectified by D3, and then passed to TR4. This transistor is connected as an.emitter follower, i.e. a current amplifier, driving MEI, a ImA meter. Resistor RI4 limits the maximum current through the transistor, preventing the meter needle from wrapping itself around the endstop if the unit is switched on with the bridge well out of balance. C3 removes ripple, giving a steady meter reading, and R12 adds a small amount of forward bias to TR4.

The function switch S2 selects either "measure" or "expose" operations. When negative measurement is required, S2c and S2d switch the balance potentiometer into the bridge circuit. Closure of S2b applies voltage to the meter circuit, and also switches on the relay driver transistor TR3 via R8, thus energising the relay RLA and turning on the enlarger lamp for focussing and measurement purposes. The diode D2 prevents voltage surges from damaging TR3 when the relay is de-energised.

## TIMER

With the function switch in the "expose" position. S2c and S2d transfer the information regarding the required exposure, contained in VR2, into the timer circuit. As $\$ 2 b$ is now open, the relay is de-energised. No supply voltage is yet applied to the timer, as RIAI and S2e contacts are open-circuit at this stage.

Contacts $S 2$ e are now closed to initiate the timing period. Resistors R2, R3 and R4 form a voltage divider, and TR2 hats a forward bias derived from the voltage developed across R4. This transistor is thus turned on, its collector volts now being close to the supply of 12 V .

The voltage dropped across R5 is applied via R6 to the base of TR3, energising the relay. The enlarger lamp is thus turned on via contacts RLA2, and contacts RLAI bypass. S2e which may now be released, as the supply voltage to the timer is now applied by the relay contacts.

## THE NEED FOR AN F.E.T.

The heart of the timer is C1, VR2 and TRI. a ficld effect transistor. The main property of a field effect transistor in this application is its extremely high input resistance. The highest value of timing resistance which can be used in any timer circuit is determined by the resistance of any circuit connected to it.

If a bipolar transistor were used for TR1, the maximum input resistance obtainable would be only a few megohms. For the timing period to be determined by the timing resistance alone, as is required for accurate peformance, the maximum value of timing resistance must be about one tenth of the input resistance of TRI, i.e. a few hundred kilohms.

As the timing period is proportional to C1 $\times$ VR2, and the timing resistance is low, CI must obviously be high. Large values of capacitance give rise to further problems, as they will have a relatively low leakage resistance, and this can also affect the accuracy of the timer. The field effect transistor, with an input resistance of several hundred megohms, allows a low value to be used for Cl , with a large value of VR2.


Fig. 2. Complete circuit diagram of the enlarger control unit

## THE TIMING OPERATION

Returning to the operation of the complete timer. TRI has a small bias applied by the voltage drop across R2. Initially CI is discharged, so the gatte of TRI is at zero volts. When the supply is connected by S2e. CI begins to charge at a rate determined by the setting of VR2. previously determined by the exposure meter. As the gate voltage approaches the bias voltage of the source, TRI begins to conduct. The gate voltage thus becomes transferred to the source by the self-biasing action of the source current. Resistor R3 also applies the source voltage to the base of TR2

When TRI source voltage, and thus TR2 base voltage, reaches approximately $1 / \mathrm{V}, \mathrm{TR} 2$ switches off. this transistor being at pnp type. The collector voltage falls to zero, switching off TR3, and deenergising RLA. The enlarger lamp is turned off, and the supply is removed from the timer by contacts RIAI, S2e having been previously released.

The standing bias, derived from the potential divider R2-R3-R4, is removed from TR1, which now becomes forward biased by the charge remaining on Cl. The capacitor is discharged rapidly through the gate-source junction of TRI and $R 2$, resetting the timer ready for the next timing operation.

## THE FUNCTION OF R1

The function of R1 is best explained by considering the circuit without this component. During the measurement operation. VR2 is disconnected from the gate of TRI. leaving Cl in a discharged condition. but voltage is applied to the timer via the contacts of RLAI which is energised.

When the function switch is changed to "expose", the timer begins a complete timing sequence, as Cl is now charged via the nowconnected VR2. The enlarger lamp will thus remain on until a complete timing sequence has elapsed. which is obviously inconvenient if the timer is set for more than a few seconds.

During measurement, RI is switched in, maintaining Cl in a fully charged condition. When the function switch is changed to the "expose" position, the timer is in the same state as at the end of at timing period, and the lamp is switched off.

The power supply for the exposure meter and timer is conventional. A full-wave bridge rectifier. D4 to D7, is fed from the mains transformer T , the resulting pulsed d.c. being smoothed by C4. The relay is then supplied via $R 9$, and the remainder of the circuits are fed from a 12 V rail stabilised by R7 and the Zener diode DI

## LAMP STABILISER

The basic principle of the stabilised supply to the enlarger lamp is illustrated in Fig. 3a, which shows two cycles of the voltage across the lamp. Only part of each positive half-cycle is applied to the bulb, although the whole of each negative half-cycle is applied.

If the supply is fed to the lamp at point $A$, the result will be a higher r.m.s. voltage for the complete cycle than if the supply were not allowed to reach the lamp until point $B$.

Thus, a rise in mains voltage can be counteracted by delaying the point in the positive half-cycle at which the bulb is switched on. Conversely, a drop in mains voltage can be corrected by an earlier switch-on, allowing more of the half-cycle to reach the bulb.

Referring to Fig. 2, each negative half-cycle is fed to the lamp via D8. The positive half-cycles are controlled by the silicon controlled rectifier CSR1.

The silicon controlled rectifier, or thyristor, is an extension of the normal silicon diode. It will under no circumstances conduct when the cathode is positive with respect to the anode, and will also block when forward biased, until a third electrode, the gate, is made positive with respect to the cathode. The thyristor will then conduct until the forward voltage is reduced to zero, even though the gate voltage is removed.

Once the thyristor turns off, it will revert to its blocking state until it is again forward biased and a gate voltage is applied. Thus CSRI can be switched on at any point during the positive half-cycle as desired, by applying a gate voltage when required. The gate voltage is derived from TR6, a unijunction transistor.

## THE UNIJUNCTION TRANSISTOR

The unijunction is a semiconductor device with two bases and an emitter. It acts as a pure resistance of 5 to 10 kilohnes with zero emitter voltage. This condition is maintained as the emitter voltage is increased, with only a few microamps of emitter current, until a criticall (peak point) level is reached.

At this point, the emitter-batse one junction breaks down, allowing a large emitter current to flow. If the emitter voltage is now removed, the device reverts to its original state, with very small ennitter current, and once more acts as a resistance between base one and base two.


## TRIGGER CIRCUIT

Referring again to Fig. 2, the diode D10, TR 5 and the alssociated circuitry will initially be ignored.

At the beginning of the positive half-cycle, CSRI is not conducting. D8 is reverse biased, and is also blocking. Thus, the full mains voltage is developed across CSRI, and is applied via R22 to the trigger circuit formed by TR6.

Initially, D1l does not conduct, until the mains voltage reaches 18 V , when the Zener will clamp at this voltage. Thereafter, the supply to the trigger circuit will remain at 18 V . the surplus voltage being dropped across R22.

As soon as the supply begins to rise, C 6 starts to charge via R19. As the supply voltage is stabilised by D11, the capacitor will always charge at the satme rate. When the charge on C6 reaches the peak point voltage of the unijunction TR6, the emitter conducts, and C $C 6$ discharges through the emitter-base one junction. This current gives a positive voltage across R21, and this voltage is applied to the gate of CSRI, switching it on.

Once the trigger circuit has supplied the gate pulse. it is no longer required until the next positive half-cycle begins. As CSRI is in parallel with the supply to the trigger circuit, and as this component has a very small voltage drop when conducting. it is apparent that the trigger circuit will be inactive once it has turned on the thyristor, and will remain so during the negative half-cycle, when D8 is conducting.

Resistor R20 is included to stabilise the working conditions of TR6 against temperature variations.

## CORRECTION OF FIRING POINT

Transistor TR 5 and the associated circuitry gives the correction of thyristor firing point necessary for stabilisation against mains variations. The matins voltage is monitored by R16, in series with the lamp and CSR1.

This resistor develops a voltage across it while CSRI is conducting, i.e. during the positive hallcycle. As this half-cycle is incomplete, being chopped to provide the correction in the lamp supply, it is obvious that the r.m.s. and average voltages across R16 will not provide a suitable reference. The peak voltage is thus used.
The waveform across R16 is applied via D9 10 C5. This capacitor is charged to the peak of the waveform, and will remain at that level unless
allowed to discharge. Three discharge paths are available. C5 cannot discharge back into the supply as it is prevented from doing so by D9.

The second path is into the base of TR5. By the use of a high gain transistor in this position, a relatively high input resistance is achieved, and C5 loses little of its charge this way.

The third possible path is through its own leakage resistance, but the loss this way is also small enough to be ignored. Thus, the voltage across C 5 is proportional to the peak voltage of the mains supply. This voltage is amplified by TR5, and a portion of the amplified voltage is tapped off from VR3, being passed by D 10 to give a standing charge on C6.

## STABILISER OPERATION

Taking the action of the stabiliser unit in its entirety, and referring to Fig. 3 b , a voltage proportional to the peak of the mains supply appears across C5. This voltage is amplified by TR5, and tapped off from the slider of VR3, then applied to C6. This gives a voltage on C6 of $V_{1}$. C6 is also charged via R19, and its voltage increases until it reaches the peak point voltage of TR6, $V_{\mathrm{p}}$. The thyristor is then turned on by the pulse across R21, and the remainder of the half-cycle is applied to the enlarger lamp.

If the mains voltage now rises, a larger voltage is developed by C 5 . This results in a reduced voltage at the collector of TR5, and a lower initial voltage on C6- $V_{2}$ in Fig. 2b. Then C6 begins to charge via R19 at the same rate as before; as the starting voltage is lower than previously, C6 takes longer to achieve $V_{p}$. The thyristor is thus turned on later in the half-cycle, compensating for the rise of mains voltage.

The peak point voltage of unijunctions varies from sample to sample, and an adjustment is necessary to compensate for this. VR3 is therefore used to adjust the d.c. level fed to C6, this being the only control necessary to set up the stabiliser.

## ALTERNATIVE SEMICONDUCTORS

Many of the components are uncritical, and alternatives can be used. This applies particularly to the semiconductors.

TRI must be an $n$-channel field effect transistor. The 2N5457 (also available under the number MPF103) has been selected because it has a lower range of characteristics compared with other types. Other f.e.t.s will work satisfactorily, but may not give the same range of times-this is not too important, as the only result will be a different setting of the paper speed control when the unit is calibrated.

TR2 is required to be a $p n p$ silicon transistor, and any such component will work satisfactorily. Similarly. TR 3 and TR4 are merely required to be $n p n$ silicon types, the only limitation being that TR3 must be able to handle the relay current -100 mA at most. TR 5 is a high gain npn silicon transistorBC109 or 2N3711 are two suitable alternatives. TR6 is a unijunction-2N2646, 2N2160 or BEN 3000 will work quite happily here. If alternatives are employed, the lead-outs must be checked, as the various types may differ from those specified.
Any Zener diodes of the required ratings are suitable. and D8 may be any silicon diode with a IA 400 V rating, or better. Similarly. CSR1 can be

## COMPONENTS

## Resistors

| R1 | $4 \cdot 7 \mathrm{k} \Omega$ | R13 | $22 \Omega$ |
| :---: | :---: | :---: | :---: |
| R2 | $6.8 \mathrm{k} \Omega$ | R14 | 5.6k $\Omega$ |
| R3 | 15k $\Omega$ | R15 | 33k $\Omega$ |
| R4 | 10k $\Omega$ |  | 3.3 22.5 W (for 150W lamp) |
| R5 | 10k $\Omega$ |  | $6.8 \Omega 2.5 \mathrm{~W}$ (for 75W lamp) |
| R6 | 10k $\Omega$ | R17 | $1 \mathrm{k} \Omega$ |
| R7 | $330 \Omega$ | R18 | $270 \Omega$ |
| R8 | $4.7 \mathrm{k} \Omega$ | R19 | 150k $\Omega$ |
| R9 | 120 S | R20 | $390 \Omega$ |
| R10 | $1 \mathrm{k} \Omega$ | R21 | $56 \Omega$ |
| R11 | 1 k , | R22 | 8.2k $\Omega 5 \mathrm{~W}$ |
| R12 | 10M $\Omega$ |  |  |

All resistors $\pm 10 \%, \frac{1}{4} \mathrm{~W}$ unless otherwise specified
Potentiometers
VR1 $25 \mathrm{k} \Omega$ lin
VR2 $2 \mathrm{M} \Omega \log$
VR3 $2.5 \mathrm{k} \Omega$ subminiature skeleton preset
Capacitors
C1 $10 \mu \mathrm{~F} 15 \mathrm{~V}$ tantalum
C2 $\quad 0.1 \mu \mathrm{~F} 250 \mathrm{~V}$ foil
C3 $50 \mu \mathrm{~F} 6 \mathrm{~V}$ electrolytic
C4 $50 \mu \mathrm{~F} 25 \mathrm{~V}$ electrolytic
C5 $25 \mu \mathrm{~F} 6 \mathrm{~V}$ electrolytic
C6 $0.1 \mu \mathrm{~F} 250 \mathrm{~V}$ foil
Semiconductors
TR1 2N5457 (or MPF103)
TR2 2N3702
TR3 2N3704
TR4 2N3704
TR5 2N2696G
TR6 TIS43
CSR1 CRS1/40
Diodes

| Diodes | 12V 400 mW Zener |
| :--- | :--- |
| D1 | OA200 ( 6 off $)$ |
| D2-7 | OA |
| D8 | 1N4004 |
| D9, 10 | OA200 (2 off) |
| D11 | 18 V 1 W Zener |

Switches
S1 2-pole on-off, toggle
S2 P.O. type lever switch $4 \mathrm{CL} / 4 \mathrm{CN}$ (see text)
Relay
RLA 500 ohm coil, 2 make contacts (one heavy duty)

Transformer
T1 Miniature mains transformer, $2 \times 12 \mathrm{~V}$ secondaries (R.S. Components)

## Miscellaneous

PCC1 ORP12
ME1 1 mA meter, type MR38P
Veroboard, 5 in $\times 3 \frac{3}{4}$ in, 0.15 in pitch, copper clad
Case, 8 in $\times 6 \mathrm{in} \times 6 \mathrm{in}$ sloping front
FS1 Fuse 5A and fuseholder
2 pointer knobs
3.5 mm jack plug and socket

4 rubber feet
Heat sink, TO5
Small plastic box
Audio screened lead
any type of adequate rating. IA 400 V . The rest of the silicon diodes, specified as OA200, may be any small signal types, such as OA202, IN914, IN916. SD19.

## RELAY AND FUNCTION SWITCH

The specification of the relay and function switch gives the minimum requirements of these components. It may prove difficult to obtain exactly the contacts required.

The relay may have a different coil resistance from the 500 ohm type specified. Any resistance down to 200 ohms can be used in the circuit as it stands: relays down to 150 ohms may be accommodated by reducing R9 to 82 ohms. Two make contacts are needed; most relays will have two or more sets of changeovers, but the unwanted contacts can, of course be ignored. One set of the contacts at least must be capable of handling 1 A at 240 V .

The function switch is a P.O. type lever keyswiten. This has three positions; when the lever is pushed in one direction from the centre. the switch locks in this position; this section requires two changeovers and two make contacts. When the lever is pushed in the other direction from centre, the switch is spring-loaded, i.e. it will return to centre when released; this section requires one make contact. As with the relay, any spare contacts can be ignored.

A suitable switch, with four changeovers in each direction, is available from several suppliers, for example Home Radio. The older type of lever keyswitch, with a cylindrical knob is also suitable. As a last resort, the function switch may be separated into two parts; a four changeover wafer switch is used for tbe "measure" part, and a separate pushbutton of the press-to-make, release-to-break type initiates the timer.

## OTHER COMPONENTS

Alternatives to the specified components must not be used in two positions. Cl must be a low leakage capacitor, tantalum being the most suitable type. Paper capacitors are available, but tend to be buiky and expensive. The more familiar aluminium electrolytic should not be used; in fact, the timer will fail to operate with such a component.

This unit requires a transformer with two separate secondaries. Most of the small mains transformers available have a single centre-tapped winding, i.z. $12 \mathrm{~V}-0-12 \mathrm{~V}$, but these should not be used. The transformer specified hats the necessary secondaries.

The value of R16 must be changed according to the size of the enlarger lamp. A 150 W bulb requires $3 \cdot 3$ ohms in this position; 6.8 ohms is used with a 75W lamp.

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Full constructional. details and how to use the Photo-Print Process Control Unit

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Start
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Cold, damp weather usually means difficult starting particularly with an ageing battery. You can remove this nagging uncertainty with this ignition booster.

## Oscilloscope beam splitter

This valuable aid to those experimenting with digital i.c. systems converts a single beam oscilloscope for viewing four digital waveforms simultaneously.

## Fibre optic light guides

The transmission of optical information from one place to another, without having to rely on line-of-sight means, gives rise to some interesting applications that do not require electrical coupling to a sensor.

## PRACTICAL <br> ELECTRONICS

FEBRUARY ISSUE ON SALE JANUARY 14

# Brerote <br> IN EIECTRONIC ENEINEERING 

By A. Freeman, B.Sc., A.R.I.C. (The Marconi Company lto)

AlThough, during the last two decades, the development of transistors and printed circuitry has led to considerable economy of space per equivalent circuit function, the requirement for more sophisticated apparatus has meant that electronic capital equipment now in use may occupy about the same volume as did the simpler systems of 20 years ago. Thus, the reduction in the consumption of metal by the electronic capital goods industry, that would have been expected as a result of the economy of space, has not taken place.

There has, however, been an increase in the use of aluminium alloy because of its lightness, workability, and better corrosion resistance than steel. For the chassis or cases in which circuits are built, cadmium plated steel is often used because of the difficulty of making solder joints in aluminium.

## RADIO TELESCOPES

Radio astronomy and communication satellites have led to the erection of large steerable reflector "dishes." Such "dishes" are made from aluminium alloy or stainless steel and must be supported so that their profile is accurately maintained, even in galeforce winds.

Cable and Wireless satellite-communication earth terminal at Bahrain with $90 f$ diameter dish antenna

Thus, there has grown up a heavy constructional industry, ancillary to the electronics side of the business, using the techniques and materials of ship and bridge-building, but with the accuracy of light engineering.

When these devices are erected in climates where winter temperatures are appreciably below freezing point, girders made from ordinary mild steel cannot be used, because at these temperatures they become brittle; comparatively light mechanical shocks can result in fracture. Therefore special steels are used which retain their toughness in very cold conditions.

## PRINTED CIRCUITS

Copper is still the most used electrical conductor, not only in the form of wire, but in increasing quantities on printed circuits, where the connections are made by etched foil on an insulating baseboard.

The copper foil for the manufacture of the clad laminates is made by electro-plating polished stainless steel strip and parting the two metals mechanically, which is not difficult because the electro-deposit has only poor adhesion to the polished surface. This method produces foil that is polished on one side and dull on the other.

Erection of the aerial for satellite communications on Ascension Island


The dull side is coated with resin and applied to the surface of the insulating laminate at the same time as fabricating the laminate. The roughness of the dull surface of the copper foil acts as an adhesion key.

Connections are made to the printed circuit either by soldering or, more generally, by sockets containing spring contacts which mate with circuit terminations on the edge of printed-circuit boards. The contacts and the circuit terminations are often goldplated to prevent tarnishing, which would otherwise lead to poor contact and "noise" in the circuit

## THIN-FILM CIRCUITS

Thin-film devices are similar to printed circuits in that all interconnections are laid down, but they also contain resistors and capacitors as integral parts of the circuit. These devices are made by evaporation techniques. The substrates are glass or ceramic; the metal to be used is evaporated in a vacuum with the metallic vapours passing through a mask to condense in the required pattern on the substrate. The substrate must be kept perfestly clean.

The materials used in this process include gold for the circuit nichrome for resistors, and silicon monoxide as insulation. Capacitors are built up by successive layers of metal and silicon monoxide. As in the case of printed circuits, the manufacture of thin-film devices readily lends itself to automation.

## WAVEGUIDES

Brass is used in the manufacture of waveguides. particularly those of large anerture, which are made from sheet as opposed to the smaller drawn tube. In areas where ferrous metals cannot be used because of electro-magnetic fields, brass screws are employed and springs are made in either phosphor-bronze or, for more severe duty. copper-beryllium.

Light alloys are widely used for waveguides, waveguide components and television camera bodies, especially the magnesium alloys N4 and N6 which have good corrosion resistance and good mechanical

Development stage of thin-film production

properties. They are very suitable for machining by methods such as routing, and N4 alloy can be welded without much difficulty.

Satisfactory brazing is really only possible with unalloyed aluminium but this has poor mechanical properties. As there is no perfectly satisfactory solder for use directly on aluminium or its alloys, the problem of the low temperature joining of aluminium alloys is generally overcome by copper-plating the alloy, tinning the areas to be soldered, removing the exposed copper chemically and finally sweating the tinned surfaces together.

When weight conservation is important, as in television camera bodies, magnesium-rich alloys are very attractive, having a density of 1.87 compared with 2.87 for the engineering alloys of aluminium. Unfortunately, these alloys readily corrode if left in an unpainted condition even in mild environments, so they are not suitable for the manufacture of items containing moving parts that cannot be painted.

## STAINLESS STEELS

There are two main types of stainless steel, austenitic and martensitic. Austenitic stainless steel is relatively soft, non-magnetic and very resistant to corrosion. This type is often used for domestic items.

Martensitic stainless steel can be hardened, is magnetic and though much more resistant to corrosion than ordinary hard steel is not as resistant to staining as the austenitic type. Table knives are made of martensitic stainless steel.

Both of these types are widely used in the electronics industry but in small quantities. The austenitic steels are particularly useful because of their non-magnetic properties. especially in the fully annealed, or mildly cold-worked condition, when they have the mechanical properties of mild steel.
Some of the special austenitic stainless steels have found use in unusually arduous conditions, e.g. type EN58J is used for equipment that is to be immersed at such a depth in the sea that oxygenation is only very slight. (Oxygenation is necessary for the prevention of corrosion in most stainless steels.)

The assembly of an r.f. module for Skynet II on the left. A completed module, with the cross-welds necessary for weight-saving, on the right


## WAVEGUIDE COMPONENTS

Waveguide components are often difficult or even impossible to make from the normal rectangular tube. In such cases, recourse is made to electroforming techniques, using stainless steel formers.

When the shape is such that a steel former could not be withdrawn at the end of the electroplating operation. a thermoplastic former of the desired shape is made. After electroforming, it is removed either by melting or by dissolving out with a suitable solvent. Electroforming methods are slow and a single component may take several days to build up to the required thickness for adequate rigidity.
A recent innovation is the production of simple waveguide components, for example. tapers. by metal spraying techniques. zinc or silver-tin being the metals most of ten used. A thin deposit about 0.010 in thick is applied to the mandrel that has previously been treated with a suitable release agent. Before removing the matndrel, the coating is strengthened with resin-glass cloth laminates wet-lad on to the sprayed metal.

In an increasing number of cases it is essential that the dimensions of waveguide components rematin constant in spite of varying temperature. This requirement is achieved by the use of alloys having a low coetficient of expansion, such as invar and nilo. The poor electrical conductivity of these alloys is overcome by electroplating with silver of sufficient thickness that all the radio frequency current is confined to the electro-deposit.

## SOLDERABILITY

Alloys similar to invar are used for the leads of transistors and miniature valves in order to obtain the satisfactory glass-to-metal seals required by these components. This application gives rise to soldering problems. because these alloys are not solderable using non-corrosive resin-based fluxes.

In most cases the leads are gold-plated in the course of the transistor manufacturing process but, because the gold is readily soluble in soft solder. the joint is in fact made to the low expansion alloy.

Removing a high-power valve, weighing about 90lb, from a broadcast transmitter



A series of isolators for use in waveguides
free copper and nichel whilst their supports are of refractory metals such as molybdenum. tantalum or tungsten sheathed in platinum.

All the metals must be of a high purity, in particular being free from gaseous impurities. Valve manufacturers experience great difficulty in obtaining refractory metals of sufficiently high purity.

## NEW APPLICATIONS

Investigation into the behaviour of metals and other substances at temperatures well removed from ambience has disclosed properties not normatly encountered, some of which are useful for specialised applications. Typical examples are the electrical super-conductivity of certain metals at temperatures near to absolute zero $\left(-273^{\circ} \mathrm{C}\right)$. It is possible that continuing research on these lines maty have repercussions on the types of metal used throughout the electronics industry

## ACKNOWLEDGEMENI

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Marconi Mark VII colour television camera at the Leeds studios of Yorkshire Television


## NEWS BRIEFS

## MICRO-COMPUTER DEVELOPMENT

The US space agency NASA, has selected RCA to develop a test model of a space computer that would be 100 times smaller and lighter than equivalent commercial systems.

The microprocessor could be the forerunner of a system for future manned and unmanned space vehicles such as the Space Shuttle and the Earth Orbiting Space Station.
The shuttle is a reusable craft that will take off like a rocket and land like an aeroplane. The space station would be a permanent orbiting laboratory able to accommodate a contingent of scientists.
Heart of the computer under development by RCA will be 15 large-scale integrated (LSI) arrays-one-eighthinch square chips each containing up to 600 electronic elements. Although the computer will weigh just 10 pounds, occupy one-half cubic foot and require only 15 watts of power, it will be capable of processing functions equivalent to room-size commercial computers.

## CHANNEL SURVEILLANCE SYSTEM

The first radar station to be ordered for the purpose of monitoring shipping movements in the Channel has been installed for the Department of Trade and Industry at H.M. Coastguard station at St Margaret's Bay, near Dover. A Decca 48 mile HR 729 system with two 16 in displays, a 9 ft scanner and 25 kW transceiver, it will range from Boulogne to Dunkirk and over a similar are of the Eastern approaches to the Dover Straits.

This is the direct outcome of the Government's concern for the accident rate in the Channel, which ied to the Nationa! Physical Laboratory with the assistance of Decca, carrying out two major traffic surveys based at Dungeness and St Margaret's Bay, for the Department of Trade and Industry.

The new radar is the first ever sold to H.M. Coastguard and will also be used for search and rescue of vessels in distress when occasions arise. Its choice follows successful trials at Deal of a smaller Decca radar which first caught the headlines when, in conditions of thick fog, the Deal Coastguard homed the Ramsgate lifeboat onto a yacht in distress and subsequently conned both back to Ramsgate Harbour.

## EVENING LECTURES

A SHORT course of six evening lectures will be held at 6.30 to $9.0 \mathrm{p} . \mathrm{m}$. All enrolment enquiries should be made to the Polytechnic
January 12 M. Hughes General Introduction to D.I.Y. Electronics

January 19 A. Douglas
January 26 F. Hyde
February 2 A. J. Dunn
February 9 A. J. Dunn
February 16 M . Hughes

Electronic Music Synthesis (including ref: to the P.E. Organ) Radio Astronomy (building a simple radio telescope) Radio Control of Models Electronics for Automobiles Logic Demonstration

All the above are regular contributors to
Practical Electronics


## WATER ON THE MOON

Till now, the examination of the rocks and dust brought back from the Moon by America and Russia, has indicated that the moon is entirely without water. Doubt is now thrown on these conclusions by Dr J. W. Freeman (chief Apollo programme investigator for the instruments) of Rice University, and a statement has been made about his reasons for deciding that there is water in pockets on the Moon.
The effiect that this will have on the future of moon stations is very considerable for it would mean that the crew would not have to bring water with them from earth, or manufacture it from chemicals or re-cycling processes.

Dr Freeman and his associate Dr Kent Hills base their statement on the fact that the instruments left on the moon by Apollos 12 and I4 detected a water cloud that was like a geyser in eruption. The vapour was very diffused and probably amounted to about a quart of water. The eruption was from a crack in the moon's surface and the event was recorded on instruments which are called suprathermal ion detectors.
The largest event that was detected covered an area of 10 square miles, and was in the region of the Sea of Storms. The accuracy of the exact position is not as good as could be hoped since there are only two instruments operating. However. the precise location can be established later.

One other point that supports the contention by Dr Freeman is that the geyser coincided with small moon quakes recorded by the seismographs left on the moon. The advent of cracks is therefore well supported.

Planetary scientists have always held that in the solar system planets all contain carbon dioxide and
water in their interiors; with the geyser like eruptions on the moon is the proof of the fact. It will of course require further supporting evidence before acceptance but no doubt this will be dealt with on the remaining two missions.

## ROCK HARVEST

The geophysical harvest of the Apollo 15 mission has been excitedly acclaimed by scientists. It will take a great deal of time to examine fully and analyse the rock and dust samples amounting to about liolbs. They have a much wider variety of types and some 60 large pieces. The rocks which have been called Genesis are not as hoped the primeval lunar material.

One of the reasons for the choice of the Fra Mauro site on the Apollo. 15 mission was the hope that primitive finds would be made. However, the amount of information that will be obtained from the samples is very important.

## HOT MOON

The moon is a hot body with the heat concentrated in the interior or alternatively in pockets of radioactive material. These pockets are perhaps around a hundred miles down from the surface.

This evidence of the heat gradient was obtained from the two probes which were put down to a depth of 6ft into the moon by the Apollo 15 astronauts. The heat flow measurements showed that the increase in temperature amounted to about one degree farenheit for each increase of one foot in depth.

Compared with the earth this is about one fifth of the escape flow. This would indicate that there are radio-active minerals inside the moon. If the abundance is evenly distributed in the body of the moon then there must be a molten core. Melting temperatures would appear at about 300 miles below the surface. If this is not the case then the radio-active materials must be in pockets at about 100 miles depth.

## MOONQUAKES

There are now on the moon three seismic stations which were set up by the crews of Apolios 12, 14 and 15. The network has brought an immense amount of new knowledge of day to day happenings on and in the moon which were not even suspected. The happenings were disguised from previous observations because of the rigidity of the shell.

The power of some of the tremors would undoubtedly have made cracks in the earth's surface which is much less rigid than the moon. The major effects take place when the earth and the moon are near to each other each month. The forces that are set up due to the increased tidal pull and the effect
of the changing barycentre cause the tremors which are transmitted back to earth. One such record showed that the disturbance originated some 500 miles below the surface which may be even greater than those occurring on earth.

The moon then is a body that is convulsed internally without visible signs on the surface.

## CORE SAMPLES

A $8 \frac{1}{2} \mathrm{ft}$ core sample was taken from the moon and this was found to contain some 57 layers of soil. This illustrates about 2.400 million years of the history of the moon. The layers are of varying thickness and range from half an inch to five inches. It is possible that each layer represents a different meteorite impact.

At the bottom of the sample core a $3 \frac{1}{2}$ in section was of much coarser material containing some chips up to $\frac{1}{2}$ in diameter. This could be basic rock which was overlaid by the rest of the material.

## MYSTERIOUS MINIOUAKES

One of the mysteries of the seismic recordings of the moon are the swarms of minor disturbances quite unrelated to tidal effects.

The majority of the activity on the moon lies in a small area below the Ocean of Storms and this is where the deep tremors are recorded. These may be due to heat from the depths of the moon. The mini swarms however, cannot come from this source.

One suggestion is that they are triggered off by the mascons. . These are areas on the moon where there is a specially dense region. These cause an out of balance effect and it may be that this is the answer. Nothing definite can be said about this at the moment. If this should not prove to be the answer then there must be some other activity that has not yet been detected.

The swarms of mini-quakes over a period of two and a half days showed four swarms and thirty events.

## SATELLITE TROUBLE

There has been some trouble with the ATS-3 satellite, whose full title is Applications Technology Satel-lite-3. It is in a synchronous orbit at 70 degrees West longitude 22,300 miles above Colombia.

Fitted on this vehicle is an aerial which should rotate at 100 $\mathrm{rev} / \mathrm{min}$. This is necessary to keep it pointed at the earth while the satellite spins. About the middle of July 1971 the aerial began to slow up and fell to $80 \mathrm{rev} / \mathrm{min}$ and sometimes stopped altogether and sometimes revolved at other speeds.

It would appear that this is due to the sun which when it is north of the equator heats up the satellite and causes the aerial to seize up. When the sun moves south of the equator all should be well again.

The block diagram, Fig. 1, shows how the huge range is encompassed in a single instrument without need for adjustment. The skin current is passed through a silicon diode which produces a voltage proportional to the logarithm of the current through it Thus equal percentage changes of current will produce equal increments of voltage, about \& millivolts for a 10 per cent change. The amplifier is in fact a differentiator below a few hertz so that an output appears only for a change of input.

## HUM AND NOISE

Anyone who has put his finger on the input of a high gain amplifier will tell you that all you get out will be an almighty hum, and possibly some Radio I as well! For this reason components are included to provide 12 dB per octave roll-off above a few hertz.

## THE READOUT

Various readouls are possible, perhaps the simplest being a voltmeter. The present instrument, however, uses a roitage controlled oscillator and a crystal microphone as a speaker, so that a lie registers as a note of rising pitch. This serves two purposes: most people can hear a small percentage change of pitch over a very wide range, so that greater sensitivity than with a meter is possible; and secondly a sort of positive feedback occurs-as the subject hears the accusing little squeak he responds by making it even worse. It is also possible to use an earpiece if the subject is to be kept unaware of his responses.


Fig. 1. Basic lie delector block diagram


Fig. 2. Circuit diagram of lie detector

## PULSE MONITOR

A bonus offered by the circuit comes in the form of a pulse monitor. If a photoconductive cell is used in place of the finger clectrodes, and the subject's fingertip placed over it, his heartbeat can be heard in the form of frequency modulation of the output tone. The way it works is that the light passing through the finger is modulated in intensity by the blood pulsing through it.

## THE 741 OPERATIONAL AMPLIFIER

If you have never used an operational amplifier before, or if you have fiddled for hours trying to get an earlier type to behave, only to find that it has succumbed to your ministrations and given up the ghost-take heart! The 741 is virtually indestructible. You can short the outputs to ground or to either supply and it doesn't mind. Even if you connect the battery the wrong way round the battery



Fig. 3. Non-inverting operational amplifier configuration
will be the first casualty (but don't try this with anything bigger than a PP3).

If you are not quite sure what an "op.amp." does it is simply this: the output goes positive if the "+" input is positive of the "-" input, and goes negative if the opposite is true. The input currents are tiny, usually less than a tenth of a microamp, and the output resistance is low, a few hundred ohms. The open loop gain-that is without feed-back-is tens of thousands as long as the inputs are not allowed to get a sniff of the output in high impedance circuits.

## CIRCUIT DETAILS

Fig. 2 shows the complete circuit diagram. An integrated circuit, $\mathbf{I C} 1$, is used as an amplifier in the standard non-inverting configuration. Fig. 3 shows this in simplified form. The gain is simply

$$
\frac{R_{\mathrm{A}}+R_{\mathrm{B}}}{R_{\mathrm{A}}} \text { or if } R_{\mathrm{B}} \geqslant R_{\mathrm{A}}, \text { simply } \frac{R_{\mathrm{B}}}{R_{\mathrm{A}}} .
$$

In the actual circuit $R_{\mathrm{B}}$ is 2.2 megohms and $R_{\mathrm{A}}$ is 470 ohms so the gain is about 4,000 at around 1 Hz . The capacitor Cl starts reducing the gain below 1 Hz , and C 2 and C 3 reduce it above $3-4 \mathrm{~Hz}$.

## COMPONENTS . . .

Resistors
R1 $2.2 \mathrm{k} \Omega 2$
R2 $5.6 \mathrm{k} \Omega$
R3 470 s.
R4 $2 \cdot 2 \mathrm{M}$ s
R5 2.2 M s
R6 68 k s
R7 33ks
R8 10ks)
R9 68ks
R10 10ks
R11 1ks
R12 2.2 k 』
R13 100k 12
R14 2.2 k S2
All $\pm 5 \%, \frac{1}{4} \mathrm{~W}$ carbon
Light Dependent Resistor
PCC1 ORP60 or U1100 Miniature CdS photocell

## Capacitors

C1 $200 \mu \mathrm{~F}$ elect. 4 V
C2 $0.015 \mu \mathrm{~F}$
C3 $0.015 \mu \mathrm{~F}$
C4 $2.5 \mu \mathrm{~F}$ elect. 64 V
Integrated Circuits
IC1, IC2 LM741CN, SN72741 or 741-OPA (8-lead D.I.P.) (2 off)

Diodes
D1 6.2 V 400 mW Zener
D2 to D6 IN914 or any small silicon signal diode (5 off)

## Speaker

TL1 Crystal mic. insert

## Switch

S1 Double pole changeover slide switch

Jack plugs and sockets
JK1 3.5 mm Jack socket with break contact, and plug
JK2 2.5 mm Jack socket with break contact, and plug

## Miscellaneous

Crystal earpiece (optional)
Plastic box 6 in $\times 4$ in
B1 9V Battery PP3
Sprung hair clips
Veroboard 0.1 in matrix, $5.7 \mathrm{in} \times 2.8 \mathrm{in}$ with copper strips
Battery connector
Capacitor clip (for mounting TL1)

The diodes D3 and D4 together with R6 to R9 reduce the gain as the output swings above zero and below -2 volts. These voltages are with reference to the positive end of the Zener. Their purpose is to reduce the time taken to recover from a very large input swing.

## VOLTAGE CONTROLLED OSCILLATOR

Fig. 4 shows the essential components of the voltage controlled oscillator. The 100 kilohm and 1 kilohm resistors apply positive feedback so that small voltages at the input cause the output to swing to its limits (about +2 and -4 volts, with respect to the positive end of the Zener). One hundredth of the output appears at the " + " input by potential divider action.

Assuming the output has just switched to its upper $(+2 V)$ limit, diode $D_{F}$ will be forward biased and the " + " input will be at +20 millivolts. Capacitor $C_{\mathrm{T}}$ will charge up through $D_{\mathrm{F}}$ and $R_{\mathrm{F}}$ until its voltage reaches the 20 millivolts on the " + " input when the output swings rapidly to -4 volts, taking the " + " input to -40 millivolts. Diode $D_{F}$ is now reverse biased and $C_{\mathrm{T}}$ can only discharge through $R_{\mathrm{T}}$ at a rate dependent on the input voltage $V_{1}$, which is assumed to be at a negative potential. As soon as the "-" input reaches -40 millivolts, the output switches to +2 volts and the cycle starts again

Thus the output is a series of pulses of almost constant width but of repetition frequency determined by $V_{c}$.


Fig. 4. Voltage controlled oscillator circuit

## THE SKIN ELECTRODES

The finger clips or skin electrodes are the only items unlikely to be found in the average constructor's inventory-they are in fact sprung hair curlers. Get the nickel-plated type rather than the aluminium ones which are difficult to solder. I got mine from Woolworths.

## MANHATTEN LAYOUT

Layout of components and cuts in the copper strips are shown in Fig. 5. The form of Veroboard layout used will probably seem wasteful at first


1234567891011121314151617181920212223242526272829303133


Fig. 5. Component assembly and wiring details of Veroboard panel (Note that the copper strip should be cut at 19T)
glance. but it hats the advantage of being casy to implement and to check. All components are more or less parallel to the copper strips, while component side links run at right angles (like the streets of Manhatten). The battery slides into three 20 s.w.g. straps on the board which prevent it from moving about, but allow it to be replaced (see Fig. 6).

Apart from the battery clips only four wires connect to the board-input, output and positive and negative supplies. The negative supply is the common rail for input and output

The miniature jacks used for input and output are of the type which break a connection when the plug is inserted.

## WARMING UP TIME

When you first turn the machine on, the large electrolytic, CI, may take a minute or so to polarise -that is, as well as taking a few seconds to charge up, and it can take considerably longer for the leakage current to reduce to its proper value. The piten of the oscillations should fall to a very low value,


Fig. 6. Wiring diagram and method of battery mounting

certainly less than a hundred hertz, before asking questions. A simple test to establish that the device is "in range" is to get the subject to take several deep breaths, this should elicit a full scale response.

## OPERATION

Put the finger clips on the subject with soothing reassurances that he won't get electric shocks. Then proceed to shock him by proving that the gadget works. A simple demonstration is to ask him to think of a number, say between one and five, and then ask him about each number in turn. He must of course say no to each number so that he must lie once. Present the numbers in random order otherwise the reaction as his number approaches might be confused with the previous one.

## WARNING

Don't try to operate the gadget from a mains supply, or battery eliminator, even small leakage currents can be felt, and larger ones could be dangerous.

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## State of the INDUSTRY

Official statistics necessarily trail events. It takes time for companies to make their returns to the Department of Trade and Industry and more time to analyse and collate them. So we had to wait until November 3 before getting the 1971 first half results on how the capital equipment sector is making out in these troubled times.

The figures were not as bad as some had feared. Compared with the first half of 1970 total output was up 5 per cent at £336.6 million. Radar was a bright feature, up 28 per cent, but slumps in broadcast equipment ( -20 per cent), navigational aids ( -11 per cent), and radio communications equipment ( -10 per cent) eliminated the gains. Control equipment put on 22 per cent and nucleonic instruments a cracking 25 per cent, although the latter is such a small segment of the industry that the gain had little effect on the overall totals. Exports were down from 40 per cent to 36 per cent. Surprisingly, employment remained constant, this year being 150,000 people at the end of both the first and second quarters.

The red light is seen in the forward order book. The figure at June 30, 1970, was £648 million. On the same date in 1971 the figure had slumped to $£ 572$ million. The conclusion must be that the great days of 15 per cent growth every year are all but over.

## LEANER AND TOUGHER

Squaring up to the commercial facts of life is no easy matter. Some people throw in the sponge as did RCA with computers. Since then the U.S. General Electric has pulled out of microcircuits. After

an investment of $£ 20$ million and four years of struggle.

Others persist by seeking economies and building better management structures. Plessey recently underwent a complete reorganisation converting a cumbersome operation into "businesses', each accountable directly to the main Plessey board.

Some details of how Mullard tackled their problem have been revealed by commercial director Jack Akerman. Mullard now has four divisions each responsible for a market sector. Thev are Consumer, Computer, Communications, and instrumentation and Control. The general manager of each division looks after his market sector not only from the sales point of view but also the Mullard manufacturing plant producing goods for his market area. Each is charged with making a profitable return on the capital employed which totals some $£ 90$ million right across the board.

Everyone is expected to work for the benefit of the whole com-pany-not just for his own division. Akerman regards the organisational change as a platform for growth in the '70s. Results after 6 months showed a reduction in stocks of over 10 per cent while providing a better service to the customer, reduction in the commercial staff and better morale.

## SLIMMING DOWN

Another example of slimming down to better meet the future as well as today's problems is to be seen in SGS whose U.K. operations are based on a marketing unit at Aylesbury and a substantial manufacturing plant at Falkirk. Fears that the Falkirk plant would be closed entirely were recently allayed by a lightning visit from Milan by Giancarlo Maimone, newly appointed managing director, who had discussions with local MP Harry Ewing and union representatives. The outcome was that Falkirk appears to be saved, for the time being at least, but with a labour force pruned from a peak of 1,100 down to some 500-600.

The marketing team at Aylesbury has also been streamlined and it is possible that a joint marketing team with ATES (a company associated with SGS through the Italian holding company IRI) will be set up. Other SGS plants in Italy, France, Germany and Singapore are all reported to be working. well under capacity. Mr Maimone has some tough decisions ahead in relation to cut-backs elsewhere but, for the moment, SGS men in the U.K. are optimistic. "At least we now have a clean situation on which to build", was the opinion of marketing manager Roy Hood.

## TELECOMMUNICATIONS BOOM

One sector of the industry which is flourishing is telecommunications which, with the introduction of PCM and electronic exchanges, is getting more and more electronic in character. The TXE 2 electronic exchange is now firmly established. Since the first was commissioned in Ambergate, Derbyshire, in 1966, some 150 have been installed and they are now coming into service at a rate of more than two every week.

The big TXE 4 with up to 40,000 lines (TXE 2 has only 2,000 lines) has now been finally ordered by the Post Office from Standard Telephones and Cables. The development of TXE 4 has not been without problems but these have now been resolved. First contract is worth $£ 15$ million and the first installation will be made in London at the beginning of 1975. Soon afterwards, TXE 4's will be in service in Manchester and Birmingham.

The TXE 2 has been a big success for Plessey who have supplied the bulk of them to the Post Office and have also been successful in selling them overseas.

But the biggest telecommunications boom area of all is data communications. Fifteen countries in Europe have now come together in sponsoring a year's study by PA International Management Consultants in London on the data communications requirements of Europe for the next 15 years. The study is planned to show future customer requirements, anticipated data traffic flow with Europe, and between Europe and other continents, and future developments in data processing systems.

## BRITAIN IN SPACE

The biggest rocket to get airborne (but only just) on Guy Fawkes day was Europa 11 launched from Kouro in Equatorial French Guiana. Regrettably, it was the biggest flop of the year.

A few days earlier the British technology satellite Prospero (formerly X3) was successfully launched in orbit by a Black Arrow from Woomera and has been functioning well.

Prospero's. success, however, will not compensate us in prestige for the loss of Europa 11. We pulled out of the European Launcher Development Organisation (ELDO) in 1968. It was a controversial move which did not endear us to our continental partners. The Blue Streak rocket used on November 5 was sold to ELDO. One crumb of comfort was that Marconi's navigational guidance system was not at fault.


## THE ARLINGTON DICTIONARY OF ELECTRONICS

Published by Arlington Books
171 pages， $8 \frac{1}{2}$ in $\times 7 \frac{1}{2}$ in．Price $£ 3$

THIS new book of American origin is well pro－ duced and attractively presented．The range of material covered is enormous，ranging from simple circuit theory to microwave engineering and quantum physics．Also included is some advanced maths including vector calculus．Digital computer fundamentals are also dealt with．Diagrams，though not profuse，are clear and well explained and the definitions are succinct and，wherever possible， assume only elementary knowledge．

My main criticism is not of this book itself but of the whole philosophy of presenting technical information in dictionary format．It is difficult to imagine when this type of book would be of use． In text books obscure terms are usually given lengthy explanations whilst a lay reader encountering technical terms would be unlikely to learn much from this type of book．

However．if this type of book appeals to you I can recommend it as an addition to library or personal book collection．

S．R．L．

## 110 INTEGRATED CIRCUIT PROJECTS FOR THE HOME CONSTRUCTOR

By R．M．Marston<br>Published by lliffe<br>128 pages， $8 \frac{1}{2}$ in $\times 5 \frac{3}{3} \mathrm{in}$ ．Price $\mathbf{5 1} \mathbf{2 0}$

IN cramming 110 circuits into as many pages Mr．Marston has undoubtedly achieved quite a feat．The integrated circuits described range from the CA3018，which is simply．four transistors on one chip．to the PA246 five watt amplifier．There is also a section describing 35 RTL digital i．c．projects．

The general procedure used is that a description of the internal circuit of the i．c．is presented，then several projects are described so that the reader becomes familiar with the function of the unit． Many of the projects are useful and interesting and it is almost certain that everyone will find something of use in this book．

The presentation of the information is，however， not all that good．Each new project does not have a heading so one has to resort to the index to find a particular circuit．All the i．c．s，apart from the digital，are represented with the operational amplifier symbol of a triangle．For i．c．s such as the CA3018 this makes it very difficult to read circuit diagrams，and on some diagrams output signals seem to come out of inputs．

At one new penn y per circuit the book is certainly good value but a little more care in presentation would have helped．

QUESTIONS AND ANSWERS ON RADIO AND TELEVISION（3rd edition）

## By H．W．Hellyer <br> Published by Newnes－Butterworth 172 pages， $6 \frac{3}{4}$ in $\times 4 \frac{3}{4}$ in．Price 60 p

THis book is clearly aimed at people who would like to know how television and radio systems operate，but whose knowledge of electronics is slight． It takes the reader through the main principles of electronics and then goes on to explain，in simple terms，the basic features of electronic devices，includ－ ing the valve and the transistor．Some of the important circuits in radio and television are then described with many circuit diagrams to illustrate practical considerations．The final chapters deal with the functions of the television receiver，including tuning，synchronising and picture linearity control． Colour reception is not dealt with in any detail．

The sections on basic electronics are lengthy and， on the whole，they provide a useful introduction to the subject for a beginner．However，in a book with this title，I think that some of the questions and answers on semiconductor manufacture would have been better replaced by explanation of some practical points，such as the methods of correct biasing for transistors．The description of the field effect transistor，which is anyway complicated by the use of the term＂depletion layer＂without definition． seems superfluous to this kind of book．

Apart from these criticisms the book provides a lot of useful information on radio and television and would give the interested beginner plenty of food for thought． S．R．L．

## RADIO AND TELEVISION YEAR BOOK 1971／72

Edited by Eric Ickinger
Published by IPC Electrical－Electronic Year Books Ltd
224 pages， $8 \frac{5}{⿱ 亠 䒑} \mathbf{i} i n \times 5 \frac{1}{2} \mathrm{in}$ ．Price $£ 1$

ACOMPLETE guide to radio and television broad－ cast stations，transmitting at v．h．f．and u．h．f．in the U．K．．，precedes the up－dated cataloguing of com－ mercial equipment under classifications：mono－ chrome television，colour television，radiograms， record players，table radios，portable radios，car radios，tape recorders，unit audio．

The format follows the usual style as a buyer＇s guide giving descriptions，brief specifications，and prices（November 1971）．Some illustrations are also included．

## HI FI YEAR BOOK 1972

## Edited by Colin Sproxton

Published by IPC Electrical－Electronic Year Books Ltd
$\mathbf{3 7 6}$ pages， $8 \frac{3}{4}$ in $\times \mathbf{6 i n}$ ．Price $\mathbf{£ 1} \mathbf{~} \mathbf{2 5}$

AMONG the preceding articles in this latest Hi Fi Year Book is an interesting outline of the Dolby noise reduction system that is now being increasingly incorporated in commercial equipment． Also included are： Hi Fi from FM radio；a short review of eighteen of the best recordings over the past year or so：the meaning and importance of pick－ up parameters；a glossary of terms associated with microphones；a list of common abbreviations and meanings．

The equipment descriptions and prices are up－ dated but do not include all new items that took their bow at the 1971 Audio Fair（see our Audio Fair report in this issue）．A complete list is given in the centre of this book of September 1971 list prices and new purchase tax rates．


## Great new CFFFE From DROTSROA M



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NPN-PNP PLASTIC
METAL CAN TYPES
Clearance of manufacturers' seconds, selected in types and guaranteed no open or short circuit manufacturers, schools and colleges.
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2N2220. BSY27-95A, BSX 44 - ${ }^{2} 6-77$. TYPE STPIB. Silicon Planar Transistors DMP
TO-IB Meral Can. Types similar to: BCY70-72, TO-18 Metal Can. Types similar to: BCY70-72, $2 \mathrm{~N} 2906-7$. 2 N 2411 and $\mathrm{BCI} 86-7$. Also used as
complementary to the above npn type device type complem
STNIB.

Price: 500 69; 1.000 E15
TYPE STN5. Silicon Planar Transistors non TO-5 Metal Can. Types similar to: BFY50-51-52 and
2N2192-92.
TYPE STPL. As above but in pnp and similar to types 2N5354-56, 2N4058-2N4061 and 2N3702-3. Also used as complementary to the above npn
devices type STNL. devices type STNL

Price: 500 E7.50; 1,000 $\mathbb{E 1 3}$ TYPE STNK. Silicon Planar Plastic Transistor npn with $10-18$ pin circular lead configuration, $1 . C$.
$200 \mathrm{~mA}, 300 \mathrm{~mW}$ and similar to BC107-8-9, BCITO. $200 \mathrm{~mA}, 300 \mathrm{~mW}$ and similar to BC1078-9,
BCI 173 , $\mathrm{BC} 182-184, \mathrm{BC} 257-8-9$ and $\mathrm{BC} 337-8$.
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30 BCl 43



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17
BCY70
$12 p$
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30p \& GET880 \& 15p \& ORP60 <br>
37p \& ORP61

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\& 30p
\end{tabular}

 \begin{tabular}{l|ll|ll|l}
$15 p$ \& BFW 10 \& 55p \& MPF102 \& 43p \& V405A <br>
20p \& BF 29 \& 27p \& MPFIO5 \& 43p \& V410A

 

205p \& BF $\times 29$ \& 27p \& MPF105 \& 43p \& V410A <br>
85 \& BF 84 \& 20p \& OC19 \& 30p \& 2G30
\end{tabular}

 $\begin{array}{llllll}\text { 75p } & \text { BF } \times 86 & \text { 22p } & \text { OC22 } & \text { 30p } & 2 \mathrm{G} 302 \\ \text { 80p } & \text { BF } \times 87 & \text { 25p } & \text { OC23 } & 33 p & 2 \mathrm{G} 30\end{array}$

 $\begin{array}{lllll}\text { 22p BFY50 } & \text { 20p } & \text { OC25 } & \text { 25p } & 2 \mathrm{G} 308 \\ \text { 45p } & \text { BFY } & \text { 20p } & \text { OC26 } & \text { 25p } \\ \text { 2G } & \text { B }\end{array}$ | 45p | BFY52 |
| :--- | :--- |
| 60p | BFY 53 | $\begin{array}{ll}\text { 60p } & \text { BFY53 } \\ \text { 70p } & \text { BSX } \\ \text { 35p } & \text { BS }\end{array}$ 35p

35
35
35
BS
35p
45P
BSY26
25p

BSY28 | 20p | OC26 | 25p | $2 G 309$ |
| :--- | :--- | :--- | :--- |
| 20p | OC28 | $40 p$ | $2 G 339$ |
| 17p | OC 29 | $40 p$ | $2 G 330 \mathrm{~A}$ | $2 \mathrm{N918}$



## Combines Mic or Tuner inputs into a single channel

By M. J. Bunce

THis article describes a simple three channel, audio mixer using an integrated circuit preamplifier. It has a substantially flat frequency response in the audio range, and is designed for use with both domestic and semi-professional tape recorders, particularly with regard to cine soundtrack recording.
The mixer can also be used by pop groups, using impedance matching transformers if low output impedánce transducers are used, such as dynamic or ribbon microphones.

## CIRCUIT DESCRIPTION

The circuit of the mixer itself is given in Fig. 1 The input lines to the pre-amplifier are blocked to d.c. by capacitors $\mathrm{C} 1, \mathrm{C} 2$ and C3. This is necessary to prevent crackling and hiss caused by current flowing through the potentiometer wiper circuits.

The pre-amplifier is an integrated circuit operational amplifier contained in a standard eight-lead TO5 package. Such amplifiers are beginning to appear quite cheaply on the surplus market.

The circuit described was designed around an S.G.S. «A709, which has an open loop gain in the order of 50,000 at frequencies up to around 1 MHz .

The gain of the amplifier for a single channel is given by:

$$
\mathrm{G}=\frac{R_{\mathrm{F}}}{R_{\mathrm{I}}}
$$

where $R_{F}$ is the value of feedback resistance R 5 in ohms and $R_{1}$ is the value of input resistance. In this case :

$$
\text { Gain } G=\frac{1,000,000}{100,000}=10
$$

Thus the amplifier has an overall gain of ten.

## CAPACITOR ROLES

To prevent the amplifier from amplifying stray radiation, picked up by the external connecting leads, the input is shunted by a resistor/capacitor network, consisting of R6 and C4. Similarly, generation of radio frequency parasitic oscillations within


Fig. 1. Circuit diagram of audio mixer

## I.C. AUDIO MIXER

## COMPONENTS . . .

Resistors
R1 $100 \mathrm{k} \Omega$
R2 $100 \mathrm{k} \Omega$
R3 100k $\Omega$
R4 15k $\Omega$
R5 $1 \mathrm{M} \Omega$
R6 $1.8 \mathrm{k} \Omega$
R7 $68 \Omega$
All $\pm 10 \%, \frac{1}{4} \mathrm{~W}$ carbon
Potentiometers
VR1 $1 \mathrm{M} \Omega$ carbon log.
VR2 1 M $\Omega 2$ carbon log.
VR3 $1 \mathrm{M} \Omega$ carbon log.

## Capacitors



C1 $4 \mu \mathrm{~F}$ elect. 15 V
C2 $4 \mu \mathrm{~F}$ elect. 15 V
C3 $4 \mu \mathrm{~F}$ elect. 15 V
C4 22 pF polystyrene
C5 47pF polystyrene
C6 $4 \mu \mathrm{~F}$ elect. 15 V
C7 $4 \mu \mathrm{~F}$ elect. 25 V
C8 $4 \mu \mathrm{~F}$ elect. 25 V
Integrated Circuit
IC1 $\mu \mathrm{A} 709$ (S.G.S.) or equivalent

## Sockets.

SK1-SK4 Standard type jack sockets (4 off)

Switch
S1 Double pole on/off switch

## Batteries

BY1, BY2 9 volts type PP9 (2 off)

## Miscellaneous

Aluminium panel 16 s.w.g. 11 in $\times 5$ in
Skirted control knobs numbered $0-10$ over 300 degrees
Veroboard 5 in $\times 2 \frac{1}{2}$ in 0.15 in matrix
Hardboard and wood as required


MAKE BREAKS TO COPPER AT:

| 4 A | 4 P | 9 F | 17 F | 30 C |
| :--- | :--- | ---: | ---: | ---: |
| 4 B | 5 F | 11 H | 18 M | 30 N |
| 4 C | 5 H | 14 G | 22 G | 300 |
| 4 N | 5 L | 14 H | 30 A | 30 P |
| 4 O | 8 H | 161 | 30 B | 32 G |

Fig. 2. Component layout and interwiring of mixer

the amplifier is suppressed by capacitor C5 at the amplifier output.

The values of these components determines the upper frequency response limit of the amplifier; its lower limit is determined by the values of the input capacitors C1, C2 and C3, and also the output capacitor C6. For this mixer application, a fairly sharp roll-off is required above 25 kHz .

Any d.c. component in the amplifier output (e.g. that due to offset or drift within the i.c.) is blocked by capacitor C6, which also protects the amplifier against short-circuit fault conditions at the input stages of the equipment to which it is connected. It may seem at first that using the component values specified voltage offsets might easily occur. In practice, however, although offsets of up to $\pm 1 \mathrm{~V}$ can occur at the output of the amplifier, this is of no consequence since the output is d.c. blocked, and the output voltage swing never exceeds $\pm 0.5 \mathrm{~V}$.

## INPUT INTERACTION

It would also appear that the inputs might interact (i.e. turning one input to minimum would cause a drop in gain of the remaining inputs). This is also not true, since the input of the i.c. is a virtual earth, and the inputs do not therefore have any effect whatsoever on each other.

As the power requirements of the amplifier are small, it was decided to use two 9 V batteries, particularly as the amplifier requires a centre-tapped


Fig. 3. Case assembly details
supply. The design of a suitable power unit for such a circuit could be disproportionately costly and complex. Besides, for the particular application of this mixer, portability is useful.
Earthing is not critical, although the fewer the number of separate earths, the better.

## CONSTRUCTION

The unit is built around the front panel, thus avoiding numerous trailing wires, so that it may be fitted into the case as a complete module. Component assembly details on the front panel are shown in Fig. 2.

The electronic components are all mounted on a piece of Veroboard which is spaced from the panel by four 6BA spacers. The integrated circuit package is mounted on an i.c. socket to obviate the risk of damage when soldering the leads.

The layout of the components on the panel is not critical, however all unused whole copper strips should be electrically bonded to earth to avoid pickup of extraneous signals.

When laying out the front panel (see photograph), consideration was given to the ergonomic aspects of the design, particularly as the cine sound track operator could well be working outdoors in cold weather. As he could be wearing gloves under such conditions, plenty of room was left between the control knobs, which are of skirted type.

The case is of thin plywood, angled to provide a comfortable degree of rake to the front panel. Since each individual constructor may wish to design his own case to suit his requirements, only a simple perspective view of the prototype case is included (Fig. 3).

## HUM SUPPRESSION

The inside of the case is lined with aluminium foil to prevent pick-up of external interference such as mains hum. The foil is earthed to the front panel at the chassis solder tag CH (Fig. 2).

The screws fitted to the front panel are all countersunk, with the exception of the four case securing screws, as this improves neatness and avoids any risk of fouling the skirt of at least one of the potentiometer knobs.

## USING THE MIXER

In setting up the mixer each number on the control knobs represents a gain increase or decrease of approximately 1 dB with the input and feedback resistance values given in Fig. 1.

For amplifier gain variations the input resistors R1, R2 and R3 can be changed either singly or totally to suit individual requirements. Thus, with the 1 Megohm feedback resistance given and input resistors in channels 1, 2 and 3 having values of 1 Megohm, 100 kilohms and 10 kilohms respectively, these will provide suitable gains for an f.m. tuner, crystal microphone and dynamic or ribbon microphone providing the latter two are fitted with impedance matching transformers.
No master gain control is fitted since for its particular design application the tape-recorder input gain control is satisfactory. Monitoring is carried out in the normal way using the recorder facility.

# - ELEGTRONORAM 

## Mowing by Moonlight

RICH. insomniac Americans can now while away the midnight hours by mowing the lawn. without fear of disturbing their neighbours.

The Elec-Trac is a quiet and compact electric tractor developed by General Electric for use in the garden. It can mow. till, blow away snow, or provide a mobile power source for electrical equipment.

A programmed starting system prevents the operator from applying too much speed too soon and when the controls are set to "full forward" a built-in safety circuit automatically accelerates the vehicle smoothly to the desired speed.

Six long-life batteries will provide eight to ten years of life under normal conditions.

The company predicts a great future for after dark lawn care. They say, "We know that from three years of testing at night that quiet mowing in the cool of the evening or early morning is a real pleasure."


## Electronic Cash Register

To meet the need for faster check-out in supermarkets and stores. a new electronic system has been designed to replace the slow and cumbersome cash register. Pitney-Bowes call the new system "SPICE" (Sales Point Information Computing Equipment), and the unit which the customer sees is closer to a computer terminal than a cash register

By connecting all the units to a central in-store controller. all information concerning sales can be recorded on magnetic tape. This can later be processed by a computer and any facts and figures which are needed can be extracted for use by the store manager. As well as accurately recording sales, the terminals may be used to get immediate notification of a customer's credit standing. The terminals can also be used as calculators to work out quantity prices or discounts.
An optional extra to SPICE is "PEPPER" (Photo-Electric Portable Probe/Reader) which eliminates the need for the sales clerk to ring up the prices. All sales items are individually marked with specal tickets, on which the price is coded into a row of bars. The electronic "pen" is quickly passed over the ticket. instantly displaying price, item number, and department number on the terminal register. Speed and elimination of human error make this very attractive to both customer and seller.

For small stores or places where access to a central computer is not feasible, a self-contained system known as "SALT" (Stand Alone Terminal) is available which can be programmed to provide on-the-spot point-of-sale information.

# A $\square \square \square \square \square$ ELECTRONC 

## A Sight for Sore Ears

When noise exceeds a certain level it is not only annoying. it can be positively dangerous! In an attempt to lessen the risk to hearing for people who live or work in noisy environments. the Noise Abatement Society is now marketing two new pieces of equipment: a survey meter which lights a lamp if sound exceeds the danger level and ear defenders for protection when it does.

## Eye Movements Control Wheelchair

For patients who have lost the use of both arms and - leys. this wheelchair offers new hope. The only actions needed to start, steer or stop the wheelchair are movements of the eyes. The sight-guide steering mechanism used in the chair was developed originally for NASA, the U.S. space agency, which is seeking ways of freeing the hands of pilots and astronauts for other in-flight duties.

A prototype wheelchair controlled by eye movements is at present being tested at the Institute of Rehabilitation Medicine in New York. If it is eventually put into production, an estimated 100,000 quadriplegics (those with no use of arms or legs) could be made mobile.


## Microstrip Production

$W^{\text {ith }}$ the increasing use of microwave communications. EMI's activities in microstrip circuitsminiature microwave devices-have rapidly expanded in the last year. Their range has grown to include doppler generators. Gunn effect power sources, aerials, amplifiers and switches.

These new microwave technioues are being applied in many areas including the development of an experimental digital radio link for the Post Office. This will be designed to transmit and receive digital signals at 120 million bits per second.

The photo shows part of the assembly process using a vacuum probe micro-manipulator.


# Use integrated circuits to simplify control functions 

# LOGICAL ーーーーローーーーーー RADIO CONTROL <br> By A．J．Dunn 

## PART TWO

CODER 2 is described in this article in two variants． Although more expensive to make than Coder 1 described last month，they are easier to assemble． Both variants use the same basic printed circuit pattern apart from a minor change to the circuitry involving IC7．

## CODER 2A

Fig． 10 shows the circuit of Coder 2 A ．The grey panel area is the part which is altered to make Coder 2B．More about this later．

The particular advantage of Coder 2 is the flexi－ bility of timing since the timing resistance combina－ tion of RI－VRI can vary between 1.5 kilohms and about 40 kilohms（ $R 1$ is fixed at 1.5 kilohms）with any value for Cl ．There is no critical lower value for the minimum signal pulse length．

The integrated circuits IC1 to IC7（Type FJK101 or 74121）are monostable multivibrators，each trig－ gering the next in turn via the R－D－C network from pin 1 （ $\bar{Q}$ output），providing the interval period of about 0.32 ms ．The resistors in these networks（ R 2 ， R4，etc．）should not exceed 200 ohms．The sync pulse

## CODER 2A



## CODER 2A

Resistors

| R1 | $1.5 \mathrm{k} \Omega$ | R5 | $1.5 \mathrm{k} \Omega$ | R9 | $1.5 \mathrm{k} \Omega$ | R13 | $22 \mathrm{k} \Omega$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | $180 \Omega$ | R6 | 180s2 | R10 | 180S | R14 | $180 \Omega$ |
| R3 | $1.5 \mathrm{k} \Omega$ | R7 | $1.5 \mathrm{k} \Omega$ | R11 | $1.5 \mathrm{k} \Omega$ |  |  |
| R4 | $180 \Omega$ | R8 | 180 S 2 | R12 | 180』 |  |  |
| All $\pm 10 \%$, $\frac{1}{8}$ |  |  |  |  |  |  |  |

## Capacitors

| C1 | $0.22 \mu \mathrm{~F}$ | C 5 | $0.22 \mu \mathrm{~F}$ | C 9 | $0.22 \mu \mathrm{~F}$ | C 13 | $1.0 \mu \mathrm{~F}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| C 2 | $2.2 \mu \mathrm{~F}$ | C 6 | $2.2 \mu \mathrm{~F}$ | C 10 | $2.2 \mu \mathrm{~F}$ | C 14 | $2.2 \mu \mathrm{~F}$ |
| C | $0.22 \mu \mathrm{~F}$ | C 7 | $0.22 \mu \mathrm{~F}$ | C 11 | $0.22 \mu \mathrm{~F}$ |  |  |
| C | $2.2 \mu \mathrm{~F}$ | C 8 | $2.2 \mu \mathrm{~F}$ | C 12 | $2.2 \mu \mathrm{~F}$ |  |  |
| All tantalum bead type |  |  |  |  |  |  |  |

## Potentiometers

VR1 to VR6 $25 k \Omega$ linear carbon ( 6 off)

## Integrated Circuits

IC1 to IC7* FJK101 or SN74121N (7 off)
IC8 FJH101 or SN7430N
D1 to D7 CG83 or OA10 (7 off)

## Miscellaneous

Fibreglass base printed circuit board and etching kit

is taken from the $\overline{\mathrm{Q}}$ output of IC7, the pulse length being 0.7 CR or approximately 14 ms .

For timing capacitors ( $\mathrm{Cl}, \mathrm{C} 3$, etc.) of $0.2 \mu \mathrm{~F}$ and control resistance of 25 kilohms in series with 1.5 kilohms, the corresponding maximum signal pulse length is approximately 5 ms .

One of the NAND gated inputs (pin 3) is connected to the common "earth" line except in the case of IC4 where an initiating switch is inserted to start the cycle of operation.
The output from the Coder is taken from IC8, a single eight-input positive logic NaND gate.

Fig. 10 (far left). Circuit diagram of Coder 2A. The area for modification to convert to Coder 2 B is indicated by the grey area

Fig. 11 (left). The printed circuit pattern (full size) for Coder 2A

Fig. 12. Component layout of Coder 2A. The area where modification is required to convert to Coder $2 B$ is shown in Fig. 16

Fig. 13 (below). Wire link connections on the copper side of the board



## CODER 2A P.C. BOARD

The pattern for the printed circuit, which should be of fibreglass base, is shown full size in Fig. 11. As with Coder 1 described last month, it is worth providing an extra blank area of board at one end for mounting; this would add about $\frac{1}{2}$ in to the length of the working pattern.
Component layout is shown in Fig. 12, while details of wire link connections are in Fig. 13.
Orientation of the integrated circuits is important; if an error is made after soldering in situ it can prove a messy job to remove the i.c. unless a special desoldering tool is available.

## CODER 2B

Coder 2B is similar to Coder 2A except it has a fixed cycle time. The integrated circuit IC7 is replaced by a dual four-input Nand gate (DTL type MC832 or 932 ) equipped with "node" inputs. The outputs from IC7 are capacitor coupled to the opposite "node" inputs, so forming a multivibrator, Fig. 14.
The output from pin 6 initiates the cycle for each period $T$ (see last month). This period can be adjusted by altering the values of the coupling capacitors C16 and C17. For small variations a series resistor-potentiometer arrangement, as in the other stages, can be connected from pin 3 to the positive line and also from pin 11 to the positive line.


* MODIFICATIONS TO MAKE CODER 2B

Integrated Circuit
IC7 Change to MC832 or 932. DTL range

## Diodes

D6 and D7 not required

## Add Resistor

R15 $180 \Omega \frac{1}{6}$ watt carbon

## Capacitors

C15, C16 $\quad 10 \mu \mathrm{~F}$ tantalum or electrolytic (2 off)


Fig. 14 (left). Modification of Coder 2A circuit from part of Fig. 10 to convert to Coder 2B

Fig. 15 (above). Printed circuit pattern (full size) for Coder 2B

Fig. 16 (right). Modified area of layout for Coder 2B


The modified printed circuit pattern is shown full size in Fig. 15, while the modified part of the component layout is shown in Fig. 16.

## DECODER

A decoder is essentially a means of translating the information transmitted in pulse code to some form of sequential mode analogous to the operation of model functions.

In the decoder described here, three dual JK flipflops make up a shift register (Fig. 17) driven by a negative going pulse, derived by feeding the input signal through two paralleled Nand gates. A retriggeragle monostable is also used to detect the signal period.

The circuit shown in Fig. 17b is used to produce the negative going pulses required by the shift register inputs to 'clock' and 'clear': this should be considered in connection with the waveform diagram Fig. 18.

The input signal (Fig. 18a) is applied to two parallel NAND gates, each with a "fan-out" of eight, producing the requisite negative going signal (Fig. 18 b) adequate to drive the $6(\times 2)$ clock input loads.

The $0.1 \mu \mathrm{~F}$ capacitor is used to slow the fast edges slightly and prevent spurious ringing effects. This signal is applied to the input of the retriggerable monostable whose output (Q) is maintained high (Fig. 18c) till after the last trigger (pulse No. 6)


Finished decoder board
dependent upon Cl and $\mathrm{RI}(2 \cdot 2 \mu \mathrm{~F}$ and $20 \mathrm{k} \Omega)$ as pulse width $\tau \simeq 0.36 R_{1} C_{1}\left(1+\frac{0.7}{R}\right) R$ being in kilohms; $C$ in pF ; and $\tau$ in nanoseconds.

## CLEAR PULSE

The positive going' edge is differentiated by C 2 and R2 (Fig. 18c), the negative going spike turning off TRI whose output is a positive pulse situated in the middle of the sync period (Fig. 18d). This pulse


Fig. 17a. Interconnection of six JK flip-flops to form a shift register for decoding


Fig. 17b. Circuit of the "clock" and "clear" pulse generator on the decoder board

COMPONENTS . . .

```
Resistors
R1 \(4.7 \mathrm{k} \Omega\)
R2 \(33 k \Omega\)
R3 \(22 \mathrm{k} \Omega\)
All \(\pm 10 \%\), \(\frac{1}{8} W\) carbon
```


## Capacitors

```
\begin{tabular}{ll}
C & \(0.1 \mu \mathrm{~F}\) ceramic disc \\
C 2 & \(2.2 \mu \mathrm{~F}\) tantalum \\
C & \(0.1 \mu \mathrm{~F}\) ceramic disc \\
C 4 & \(47 \mu \mathrm{~F}\) elect.
\end{tabular}
```


## Integrated Circuits

IC1 MC846 or 946 quad 2-input gate (DTL)
IC2, IC3, IC4 FJJ121 or 7473N Dual JK (TTL) (3 off)
| IC5 F9601 retriggerable monostable
Transistor and Diode
TR1 2N3704
D1 OA202

## Miscellaneous

Fibreglass printed circuit board and etching kit
is applied to two parallel Nand gates whose negative going output (Fig. 18e) drives the "clear" inputs of the register.

The action of the "clear" pulse is to set the outputs (Q) of all six flip-flops to " 0 " awaiting the arrival of the first signal pulse which causes the output of the first flip-flop to go high' (1) (since it takes an input from an inverted output ( $\mathbf{Q}$ ) for the period of the pulse and the interval period. The second signal pulse causes the second output to change to " 1 " and the first output to revert to " 0 ".

Subsequent pulses appear individually as positive output pulses at the five flip-flop outputs.

The output from Q1 is shown at " $B$ " in Fig. $19 b$.

## SIXTH CHANNEL

The sixth channel output is obtained, if required, by the use of a further quadruple NAND gate wired as shown in Fig. 19a and if desired, associated with a servo unit. The positive going input signal at "A" is gated with the output of the flip-flop output Q1 to produce the signal shown in " C ". This is again gated with the input to produce a negative going pulse as at " $D$ ". which is inverted, using another gate to produce the sixth positive going signal output at "E"

The shift register is formed by connecting the dual flip-flops in cascade as shown in Fig. 17a, it being noted that one interconnection pair is reversed to store the necessary " 1 ".

Fig. 20 shows the printed circuit layout, the components being assembled as shown in Fig. 21. Note again the orientation of the integrated circuits and polarity of the capacitors and diode.

When carefully soldered, thin insulated wire should be used to wire up the signal connection pads. the positive supply connections and the earth connections (Fig. 22).

## DECODER VARIATIONS

The shift register can be extended for as many outputs (channels) as required provided that the "fan out" performance of parallel gates exceeds the number of loads-two per "clock" and "clear" input.

The present cost of SN7473N i.c.s invites the use of a cheaper gate to obtain the last output (Fig. 19) but a single flip-flop (FJJ111) could be used to store the " 1 ".

If Coder 2B (with positive sync pulse) is used the circuit in Fig. 17b is modified and the inverted ( $\overline{\mathrm{Q}}$ ) output from pin 6 of IC5 is applied to the base of TR1 and an output is taken from its collector (collector load $5 \cdot 1 \mathrm{k} \Omega$ ) direct to the "clear" inputs.

The "clear" input signal (which overrides the action of "clock" pulses) is maintained to the end of the cycle.

A $1,000 \mathrm{pF}$ capacitor should be connected between the clock inputs and ground ( 0 V ) to prevent spurjous ringing and the "extra" flip-flop method should be employed.

## CONSTRUCTION NOTES

It cannot be over-emphasised that care should be taken with construction, particularly the soldering: if sound components are used and the circuit wired correctly, no difficulty may be expected; however, a simple solder bridge between copper pads can cause



Fig. 19a. Quad 2-input NAND gate is used to obtain the sixth channel from the Q1 output. Waveforms are given below


Fig. 19b. Waveforms at input and outputs of the gates
Fig. 18a. A typical input to first two NAND gates in IC1
Fig. 18b. Output of first two NAND gates fed into the clock line and into the signal period detector multivibrator IC5

Fig. 18c. Output $Q$ from the resettable monostable multivibrator IC5

Fig. 18d. Output waveform from TR1 indicating the positive pulse in the middle of the sync period

Fig. 18e. The pulse in Fig. 18d is inverted by two more paralleled NAND gates to produce a sharp "clear" pulse
complete malfunction, necessitating the use of an oscilloscope to locate the fault.

The following notes may be of value:

1. The i.c.s should be obtained from a reliable source and properly identified. It is difficult to remove a "dud" and the copper printing may be stripped with excess heat.
2. The capacitors should be inspected for leaking and loose ends and then should be polarised before assembly. A "working" voltage should be applied from a battery via a current meter noting that the leakage current falls to a low value: the absence of an initial charging current indicates an open circuit capacitor and an unpolarised or short circuit capacitor will cause excess current causing i.c. failure. Tantalum capacitors or long life electrolytics are preferred.
3. The wire ends of the diodes should be carefully bent with wiring pliers to avoid fracture of the glass seals.
4. Soldering should not be attempted without a clean small bit.
5. The recommended technique for insulating wire connections is to bare the ends of an excess length of multi-stranded wire, twist up and solder the ends, causing the insulation to creep back beyond the point where baring took place. The ends are then cropped till in in of clean soldered wire is exposed: the iron is applied to the tinned copper pad until it runs freely; then the wire is applied.

## TESTING

The components should be inserted on the board and wired up as indicated, noting and checking the
polarity of each component as fitted. Before any interconnecting wires are fitted the board should be held to the light and the component connections checked against the theoretical circuit. The variable resistors (coders) may be replaced by a selection of fixed resistors (as shown in photograph) within the VR range and the current supply voltage applied, noting the current to be less than 150 mA . If an oscilloscope is available the waveform can now be examined at the output, otherwise an audio test may be applied (capacitor and headphones) noting the output of the multivibrator, each stage and at final output. The coder can now be connected to the decoder as in Fig. 1 noting that the supply current to the decoder is less than 50 mA .

An oscilloscope triggered from the coder output should show the pulse train input and individual decoder output pulses.

A d.c. meter (of $20 \mathrm{k} \Omega \Omega$ per volt resistance or better) should indicate different voltages of the order of $\frac{1}{2}$ to $\frac{3}{4} \mathrm{~V}$ at each decoder output in accordance with the values of coder resistance used. The meter performing an integrating action on the 5 V pulses. being present, say, for 4 milliseconds every 30 or $4 / 30 \times 5 \mathrm{~V}$. The variable resistors may now be fitted and their action noted in changing the decoded pulse length or d.c. output.
Note: In Part 1 last month the following corrections are necessary:
Fig. 5. Capacitor C15 should be reversed
Fig. 7. The blob of ink at bottom centre is not part of the printed circuit pattern
Table 1. The quad 2-input NAND gate by STC-ITT should be MIC9465D


## LOOK FOR THE FOLLOW-UP ARTICLE MODEL SERVO CONTROL

IN NEXT MONTH'S ISSUE

Fig. 20 (left). Printed circuit pattern for the decoder
Fig. 21. Component layout of the decoder board
Fig. 22 (below). Link wires are soldered on to the copper side for power supplies and signal lines



A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.
This is YOUR page and any idea published will be awarded payment according to its merits.

## STABILISED SUPPLY WITH OVERLOAD PROTECTION

The circuit diagram Fig. I was designed to power a 12 watt stereo amplifier. It was found to be desirable to have a constant voltage regardless of power being delivered by the amplifier. The circuit in Fig. 1 was evolved to achieve these requirements.

The potential at point $A$ is always a fixed fraction of $V_{0}$ and is determined by R2 and R3. As $R_{1}$, is lowered $V_{n}$ will lower causing the potential at $A$ to decrease. As the base potential of TR1 is fixed by the Zener and VR1 so TRI's base-emitter diode D4 becomes more forward biased. Hence more current flows into the Darlington pair TR2 and TR3 which supplies $R_{\mathrm{L}}$ with a greater current, restoring the original output voltage.


Fig. 1. Circuit diagram of the stabilised supply overload protection

Components VR2, D2 and D3 provide a current limit which is set by adjustment of VR2. Diodes D2 and D3 should be silicon types and have turn-on voltages of 0.6 V each. Transistor TR3 is a germanium transistor and so $V_{\text {be }}$ will be about 0.2 V while $V_{\text {lee }}$ for TR2 will be about 0.6 V .

When a small current is being supplied to the load the voltage drop across D2 and D3 is not large enough to affect the circuit, but if $R_{\mathrm{L}}$ becomes very small the voltage drop across VR2 is arranged to bring D2 and D3 into conduction, preventing an excessive current through TR3. A sharper limiting action could be achieved by replacing D2 and D3 with a zener and increasing the value of VR2.

The output voltage is set to the required value by VR1. To obtain low voltages, say 10 V , a bleeder resistor should be inserted across the output as TR3 has a leakage current of about 4 mA . Capacitor Cl is included so that the output voltage rises slowly when the circuit is switched on. Diode D1 is included to protect TR1 when the circuit is switched off.

The output resistance of the circuit has been calculated and is given by:

$$
R_{\mathrm{o}}=\frac{\mathrm{R} 2+\mathrm{R} 3}{\mathrm{~h}_{\mathrm{FE}(\mathrm{TR} 2)} \times \mathrm{h}_{\mathrm{FE}(\mathrm{TR} 3)}}
$$

This is $3 \Omega$ at the worst and about $0.5 \Omega$ at the best.
J. Welch,

Northampton.

## RESETTABLE COUNTER *



Fig. 1. Exploded view of the counter

1HAVE converted an ordinary electromagnetic counter into a resettable counter by removing the spindle that supports the number wheels, and replacing it by one about $\frac{1}{2}$ in longer. On the protruding end I fixed a small knob. A slot was cut in the metal cover to enable it to be slid over the protruding spindle, see Fig. 1. The counter may now be reset by turning the knob.

With some counters a second ratchet under the number wheels prevents them being reversed. A small lever may be fixed to the second ratchet to enable it to be disengaged when reversing the counter.

A. Hartley,<br>Bolton,<br>Lancs.

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# Audio Fair 

A brief look at some recent developments in the hif if industry By M. A. COLWELL

AUDIO by definition means pertaining to sounds, so perhaps one could be forgiven for expecting noises resembling several kinds of phenomena. And since the Audio Fair has been described as an exhibition of home entertainment, perhaps we could likewise expect to see an assortment of musical instruments, television and radio, or-dare we sug-gest-electronic games. However, television was strictly banned by the organisers on the day of opening, so how far should we define "home entertainment" in terms of electronics?

Today’s Audio Festival and Fair is a far cry from that high fidelity social function erstwhile held in luxury carpeted surroundings at the Hotel Russel and previously at the Waldorf.

## LONDON OR BRIGHTON?

Because of overcrowding, the Audio Fair organisers were probably right in finding an alternative venue. Consequently in 1969 we heard some convincing arguments that Olympia was suitable; that acoustically treated demonstration rooms could be successful.

Give them their due, Rex Hassan and company tried hard amid tremendous criticism, but a great deal can be learned by closer examination of the Fair as a whole and its venue. and it is quite obvious that some more hard thinking must be done to rate this exhibition as near a success as can be achieved.

Now is the right time to take stock of such irritating discomforts as would not be tolerated in any hotel premises. Among these, I place catering at the head of the list, litter strewn gangways second, and poor ventilation third. And oh! that dreadful train service!

In the previous week I sampled the exhibition at Brighton called Inter/Nepcon and subsequently conclude that these premises must supersede Olympia by decades. Apart from the sweet ozone and spotless terraced houses nearby, the Metropole has every reason to be proud of its conference and exhibition centre and I recommend it wholeheartedly to organisers of such as the size of the Audio Fair.

## BROAD RANGE

The range of exhibits at this year's Fair has broadened to include medium fi, low fi, and even (if it is possible) no fi at all. The extension of range to include electro-musical gimmicks, record and tape shops, nameplates, and the not-so-high-as-it-used-to-be-fi inevitably lowers the general image that we have become so accustomed to in the past.

The overall technical standard of equipment and products shown was generally interfered with by infiltration of medium fi, but well established high quality products were still to be found.

## RANK INFLUENCE

One interesting point to emerge is the influence of big business on the market through an established small company name. I am referring in particular
to the Rank Organisation whose acquisition of Wharfedale and Leak has already been geared towards mass marketing.

Although Harold Leak (now retired) did produce most of the parts that make up a complete hi fi system. he did not market a transcription unit. Logical first step then for Rank was to fill this gap, but not from their own stable. They have contracted with the Lenco Company of Switzerland to sell transcription units that they call the "Delta". Believe it or not, these are identical to the established Goldring Lenco GL75.

At least Rank recognises quality when it comes to such deals but it is a smack in the eye for the reputation of the former all-British Leak company. A Shure M75 magnetic cartridge is fitted to the Lenco pick-up arm, which has the usual cueing.

## NEW SPEAKER SYSTEMS

Another îrst timer from Ferrograph, who previously preferred to specialise mainly in the high quality tape recording field, is a loudspeaker system. It is a three-unit assembly designed in collaboration with a university acoustics team, and uses two Goodmans treble units and one KEF B139 woofer.

Ferrograph state that their sophisticated crossover network eliminates what little coloration exists outside the pass bands. From this we can deduce that bandpass cut-off is very sharp so that interaction between speaker units is minimal. They quote a "sensibly flat" response from 30 Hz to 15 kHz .

Metrosound Audio Products Ltd. also announces a new speaker, the Duplex 25, to match their new ST60 amplifier (or any other make). With one conventional cone type unit for the low frequency end, they have incorporated a new electrostatic unit based on a new patent specification. High frequency dispersion is said to be improved by mounting this behind a lenticular grille.

The Duplex 25 is $27 \frac{1}{2}$ in high and can handle 25 watts. Nominal impedance is 8 ohms and frequency range is quoted without reference level as 40 Hz to 18 kHz . Price at time of going to press $£ 52$.

## HI FI IN KIT FORM

Looking for further outlets for their existing speaker kits, EMI Sound Products Ltd. are now marketing a range of enclosures in kit form from $£ 5.80$ for a compact bookshelf model to $£ 29.50$ for a large floor standing model. A hammer and screwdriver are the only tools needed to assemble these; the baffle-boards are already cut to suit the appropriate kits.

Heath (Gloucester) Ltd., who sell the famous Heathkits, announce a new tuner-amplifier kit, the AR-2000. The tuner processes a.m. (L.W., M.W., and S.W.) and f.m. mono and stereo broadcast signals, using up-to-date techniques, including f.e.t front end and integrated circuits for the f.m. i.f. and decoder stages. Ceramic filters are aiso used.

The kit price at $£ 89.90$ is competitive with equivalent commercial ready made units, but is still a little expensive for a kit.

Another well established name in hi fi circles is looking to expansion at a rapid rate. No less than eight new products were shown on the Goodmans stand.

Well known for its wide range of quality loudspeakers, they are competing forcefully with the likes of the Rank Organisation. The new items include an attractive a.m./f.m. tuner amplifier the Model One-Ten at $£ 135$; the Module 80 Compact, a combined f.m. stereo tuner amplifier of high performance with record player type GL75 at $£ 165$; the Dimen-sion-8 bi-directional speaker at $£ 69$ teak, $£ 75$ white finish.

Goodmans also introduced a speaker kit, the DIN20 incorporating an 8 in bass unit and an h.f. dome radiator with a crossover unit. This is a budget price kit of excellent quality for those with limited financial resources at $£ 12.54$ including purchase tax. Goodmans also expect to, release their new professional stereo tape recorder in 1972 at a price in the region of $£ 400$. The prototype shown at the Fair looks very impressive and uses slider controls, twin capstans, and switchable 2 -speed NAB or DIN equalisation.


Duplex 25 electrostatic and cone speaker from Metrosound


Metrosound ST60 amplifier


Heathkit tuner amplifier type AR-2000


## BELT DRIVE

Past experiments on belt drive turn-tables have met with cynical comment, but progress to date is brought to fruition in the GL85 integrated transcription turntable unit and pick-up arm from Goldring.

The turntable is driven by a "non-stretch" flat belt from a low speed 16 -pole synchronous motor. Motor speed is accurately controlled by an electronic stabiliser circuit. Two speeds are available and an illuminated stroboscope indicates speed accuracy.

An electronic sensing device raises the pick-up arm at the end of the record and switches off the motor. This unit is expected to be available in January and is to be supplied complete with plinth for $£ 72.78$ including purchase tax.
The Connoisseur BD2 turntable at $£ 32$ now has push button speed change.

## STEREO PHONES

Among headphones on show, which seem now io be a common occurrence for demonstration at the Fair, Koss were dominant with the excellent PRO4AA stereo phones popping up on several stands, not the least being our own. The first four-channel headphones were also demonstrated by Tape Music Distributors Ltd. the distributors for Koss. They are the $\mathrm{K}_{2}+2$, using four separate dynamic drivers, and are likely to sell for around $£ 45$.
A.K.G. introduced the K180 (at $£ 30$ ) high quality headset with balancing volume controls.

Wilmex were demonstrating the Japanese Stax SR3 electrostratic headset with their power driver switch-over unit type SRD5. At 420 grams these are comparatively lightweight. The SRD 5 will power two pairs of SR3 phones. A complete set of one pair of phones and one power unit costs $£ 45 \cdot 40$; a second pair of phones would cost $£ 31$. In spite of reassurances I don't particularly care for high voltages so close to my ears.

## DEMONSTRATIONS

The Hi-Fi Theatre is an excellent way of getting to know people and several personalities in the world of audio engineering, with a selection of magazine writers, gave some fascinating demonstrations, albeit, we are pleased to record, without commercial prejudice.

This is an essential alternative to an otherwise biased demonstration that may be set up on manufacturers' stands. The range was very wide and we were pleased to note the "free-for-all" session on the Friday where various magazines collaborated to help the public audience with their hi fi problems. This session was later edited and broadcast on BBC Radio London in November.

## WE WERE THERE

What of Practical Electronics? Judging by the sales of P.E. Gemini Amplifier reprints, especially at the exhibition, this project is still a very popular do-it-yourself amplifier design.

The amplifier, originally shown last year, was joined this year by its new companion the P.E. Gemini Tuner, a stereo f.m. tuner of compatible design, with integrated circuit phase lock loop decoder and high selectivity ceramic filters.

Full details of this new design will appear in Practical Electronics very soon.

Also represented on the P.E. stand was our new companion Everyday Electronics, which showed among other items, the integrated circuit record player, that appeared in the November issue.

Finally, a note for your new 1972 diary; the other London audio exhibition, Sonex at Skyway Hotel, near London Airport, will be open to the public March 24 to 26 inclusive, preceded by two "trade only" days.

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| 1840 K 10 | 175p | 2N3055 |  | 40 | 55p | 24 | 24p | BCI84L | $11 p$ | BFY90 | 104p | OA95 | $6 p$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 N 914 | 5p | 2 N 3325 | 53 p | 40362 | 68 p | AFI25 | 24p | BCI86 | 42p | B5 $\times 20$ | 16 p | OA200 | p |
| 1N316 | 10p | 2 N 3405 | 60p | 40406 | 75p | AFI26 | 22p | BC212L | $16 p$ | BY164 | 45p | OA202 | 0 p |
| IN1763A | 24p | 2N3663 | 52p | 40408 | 70p | AF127 | 22p | BC213L | 16 p | BY238 | 18 p | OCI9 | 50p |
| IN3754 | 20p | 2 N 3702 | 13p | 40412 | $67 p$ | AF139 | 33 p | BC214L | $16 p$ | BYX38. |  | $\bigcirc \mathrm{OC} 25$ | 42p |
| IN5399 | $21 p$ | 2N3703 | 13 p | 40430 | 140p | AF239 | $36 p$ | BC257 | 9 p | 300 | 38p | OC28 | 70p |
| 1 N 5402 | 28p | 2N3704 | 13 p | 40432 | 185p | AL102 | 77 p | BC258 | 8 p | BY×38- |  | OC29 | 76p |
| IN5407 | 45p | 2N3705 | $13 p$ | 40512 | 195p | ASY26 | 27p | BC259 | 9 p | 300R | 38p | OC35 | 60p |
| 1544 | 9 p | 2N3706 | 13 p | 40602 | 52p | ASY27 | $36 p$ | BC267 | 17p | C407 | 17 p | OC36 | 65p |
| 15940 | 5 p | 2N3707 | $13 p$ | 40669 | 140p | ASY28 | 27p | BC268 | $15 p$ | C762 | 19p | OC41 | 42p |
| 2N696 | 17p | 2 N 3708 | 10p | AC107 | 46p | ASY29 | $36 p$ | BC269 | 17 p | Cl 412 | 102p | $\mathrm{OC}_{2}$ | 46p |
| $2 N 697$ | $18 p$ | 2N3709 | $11 p$ | AC126 | 20p | AUll | 97p | BC300 | 49p | E2512 | 164p | OC44 | 42p |
| 2N706 | 12 p | $2 N 3710$ | $13 p$ | ACl27 | 20p | B30C250 | 24p | BC301 | $37 p$ | EA 403 | 10p | OC45 | 38p |
| 2 N 930 | 29p | 2N3711 | $13 p$ | ACl28 | 20p | B30C550/ |  | BC303 | 60 p | EB383 | 10p | OC70 | $21 p$ |
| $2 \mathrm{NII31}$ | 29p | 2N3731 | 120p | ACl41H | 34p | 300 | 34p | BCY30 | 60 p | EC401 | 18 p | OC71 | 38p |
| 2N1132 | 29p | 2N3794 | 15p | ACI4IHK | 37p | B1912 | 66 p | BCY31 | 75p | EC402 | 17p | OC72 | 38p |
| 2N1302 | 19p | $2 N 3819$ | 23p | AC142H | 25p | 85041 | 72p | BCY70 | $18 p$ | ER900 | 54 p | OC75 | 40p |
| 2 N 1303 | 19p | 2N3820 | 53 p | ACl42HK | 29p | BA102 | 25p | BCY71 | 33p | MC140 | 25p | OC81 | $25 p$ |
| 2 Nl 304 | 26p | $2 N 3904$ | $35 p$ | AC153K | 22p | BA130 | $22 p$ | BCY72 | 15p | M1481 | 120p | OC8ID | 25p |
| 2N1305 | 26p | 2N3906 | ${ }^{35}$ | ACI76 | $16 p$ | BA145 | 27p | BD121 | 105p | MJ491 | 135p | OC83 | $25 p$ |
| 2N1306 | 33 p | 2N4036 | 55p | ACI76K | $17 p$ | BAl55 | 15p | BD123 | 105p | M1371 | 108p | OC84 | 25p |
| 2N1307 | 33 p | 2N4058 | 13 p | AC187K | $17 p$ | BAl56 | 13p | BDI24 | 100 p | MJE521 | 92p | P346A | 26p |
| 2N1308 | $36 p$ | 2 N 4059 | 10p | ACI88K | 23p | BAXI3 | 13 p | BDI30 | 50p | MJE2955 | 165p | S2CNI | 10p |
| 2N1309 | $36 p$ | 2N4060 | $11 p$ | *ACI87K/ |  | B8103/B | 16 p | BDI31 | 79p | MJE3055 | 82p | SCI41D | 187p |
| 2N1596 | 102p | 2 N 4061 | $11 p$ | 188K | 40p | BB103/G | 16p | BD132 | 86p | MPF102 | 37p | SC146D | 247p |
| 2N1599 | 122p | $2 N 4062$ | 12p | $A C Y 17$ | $31 p$ | BC107 | 12 p | BD135 | 38p | MPS653! | $35 p$ | SDI | 10p |
| 2 N 613 | ${ }^{23} \mathrm{p}$ | 2N4124 | 18 p | ACY18 | 19p | BC108 | $11 p$ | BDI36 | 44p | MPS6534 | 30 p | SD4 | 12p |
| 2NI711 | 26p | 2N4126 | 27p | ACYI9 | 23p | BC109 | 12p | BDI41 | 227p | NKT211 | 25p | $\checkmark 763$ | 28p |
| 2 N 1893 | 54p | 2N4284 | $15 p$ | ACY20 | 20p | BC122 | $21 p$ | BDY20 | 92p | NKT2:2 | 25p | W106BI | 45p |
| 2N2147 | 95p | 2 N 4286 | 15 p | ACY21 | $21 p$ | BC. 125 | 15 p | BFII5 | 23 p | NKT213 | 25p | Wio6DI | 83 p |
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| 2 N 2218 A | 44 p | 2 N 4291 | 15p | ACY39 | 63 p | BC140 | 30p | BF173 | 19p | NKT217 | 50p | WPO2 | 95p |
| 2N2219 | 38p | 2N4292 | 15p | ACY40 | 17 F | BC147 | 10p | BF177 | 25p | NKT261 | $21 p$ | ZT×300 | 14p |
| 2N2219A | 53p | 2N4410 | 24p | ACY4I | $18 p$ | BC148 | 9 p | BF178 | $31 p$ | NKT271 | 18p | ZTX301 | $16 p$ |
| 2 N 2270 | 62p | $2 \mathrm{~N}_{4} 443$ | $111 p$ | ACY44 | $31 p$ | BC149 | 10p | BF/94 | 14p | NKT274 | 18p | ZTX302 | 22p |
| 2N2369A | 19p | 2N4906 | 305p | ADI40 | 63 p | BCI53 | 19p | BF195 | 15 p | NKT275 | 23p | $2 T \times 303$ | 22p |
| 2 N 2483 | 35p | 2N4915 | 215p | ADI42 | 50p | BC154 | 20p | BF244 | 30p | NKT403 | 65p | Z $7 \times 304$ | 27p |
| ${ }_{2} \mathrm{~N}_{2} 484$ | 42p | 2 N 4991 | 62p | ADI49 | 58p | BC157 | 12 p | BF254 | 14p | NKT404 | $61 p$ | 2TX330 | 23p |
| 2N2646 | 47 p | 2N5062 | $61 p$ | ADI50 | 50p | BC158 | $11 p$ | BF255 | 15p | NKT405 | 79p | 2TX331 | 27p |
| 2 N 2904 | 38p | 2N5088 | 38p | ADI61 | 33p | BC159 | 12 p | BFX18 | 90p | NKT603F | 30p | ZTX500 | 18p |
| 2N2904A | 42p | 2N5163 | 25p | ADI62 | $36 p$ | BC167 | $11 p$ | BF×29 | $31 p$ | NKT613F | 30p | $2 \mathrm{~T} \times 501$ | $21 p$ |
| 2 N 2905 | 44p | 2N5172 | 18p | *AD161/ |  | BC168 | 10p | BFX84 | 25p | NKT674F | 24p | ZTX502 | $25 p$ |
| 2N2905A | 47 p | 2 N 5192 | 125p | 162 | 60p | BC169 | $11 p$ | BFX85 | 32p | NKT677F | 22p | ZTX503 | 22p |
| 2 N 2924 | 20p | 2 N 5195 | 147p | AFII4 | 24p | BC177 | $14 p$ | $8 \mathrm{~F} \times 87$ | 29p | NKT713 | 30p | Z $T \times 504$ | 52 p |
| 2 N 2925 | 22p | $2 N 5457$ | 49p | AFIIS | $24 p$ | BC178 | 13 p | BFX88 | 26p | NKT773 | 25p | ZTX530 | 27p |
| 2N2926 2N3053 | $11 p$ | 2N5459 | 49p | AFII6 | 22 p | BC 179 | $14 p$ | BFX50 | 23p | OA47 | 8 p | ZTX531 | 33p |
| 2N3054 | 27p | 4025 4 | 71p 89 | AFl17 AFII | 22p | BC182L BCI83L | $11 p$ $10 p$ | BFY5 BFY52 | 20p | OA90 OA91 | 6p | at |  |

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| $C$ | $1 / 2 W$ | $10 \%$ | $47 \Omega-10 M \Omega$ | $E 12$ |
| $M O$ | $2 \%$ | $10 \Omega-1 M \Omega$ | $E 24$ |  |
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| Outputs are continuouslyvariable, except CU600F with output in IV switched steps |  |  |

Prices start at $£ 12.00$ per unit, subject to educational

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$\qquad$


# PNTENTI RESU|ETWoos. 

IMPROVING AMPLIFIER PERFORMANCE

| N their new British Patent 1232 324, Redifon Limited discuss multistage transistor amplitiers of the push-pull lype, capable of operating over a wide frequency range with coupling of the stages directly and by an inductive device with a view to improving performance.

## BP 1232324



Redifon explain that it is already known that a pair of transistors can be biased into class A conditions and coupled by way of a transformer so as to drive a further pair of transistors biased into class B conditions. It is also known that a pair of transistors can be biased into class B conditions and directly coupled to a further pair of transistors biased also into class $B$.

But, according to Redifon, in the transformer coupled arrangement there is difficulty in providing a driver transformer for the class $B$ stage with a sufficiently low leakage reactance to transfer power over a wide frequency band with litile distortion.

Various ways round problems like these have been suggested, but Redifon claim they have a novel idea in suggesting an amplifier with a first stage in class $A$ and a second stage in class $B$, using an inductive device with a pair of tapped windings, arranged so that an output electrode of a transistor in the first branch of the first stage is connected to one end of one of the tapped windings, and an output electrode of a transistor of the second branch of the first stage is connected to one end of the other of these windings.

An input of a transistor in the first branch of the second stage is
connected to the tapping of the first tapped winding and an input electrode of a transistor in the second branch of the second stage is connected to the tapping of the other one of the tapped windings.

The two windings of the inductive device are connected in a sense such that power for driving each transistor of the second stage is provided by both the transistors of the first stage.

Redifon give a circuit diagram for an amplifier of this type and the specification also carries some specific component values. Because of the load sharing by the transistors of the first stage, regulation is improved and distortion due to the effect of intermittent load imposed by each second stage transistor, is reduced.

## SYNTHETIC ANIMAL SOUNDS

| N the name of Santa Rita Technology Inc. of California, BP 1228405 is particularly interesting because of its possible application to the projected airport at Foulness.

As is well known it is often necessary to rid airport runways of birds. This can be done by generating alarm signals, but, as Santa Rita explain, recording natural animal sounds and replaying them over a public address system is quite often only shortlived in its effect. The birds in
some strange way seem to get used to the recording and recognise it as phoney.
Their new invention is a system for producing synthetic animal sounds and in its simplest form it comprises a signal generator for generating carrier signals in the frequency range 500 Hz to $5,000 \mathrm{~Hz}$. The selected carrier is amplitude modulated by another tone in the range 50 Hz to 300 Hz and a straightforward electro-acoustic transducer is used to convert the modulated carrier signal into a sound signal.

The sound generator can be arranged to turn the modulated carrier signal on and off at a rate of one to five times per second either sharply or gradually. And it may also have a dither arrangement for dithering the frequency of the modulating means over a two-to-one range. Moreover, the carrier signal can also be dithered over a frequency range comparable to the frequency of modulation.

Consequently, it is apparently easy to vary the sound signal components and so avoid the birds getting used to any one signal.

The specification shows a block diagram of the basic layout. Some waveforms show how the "chirps" produced can be either in square wave or sinusoidal form.
Santa Rita claim special success in the diversion of red-winged blackbirds from a landing strip for jets with a $3,000 \mathrm{~Hz}$ carrier and a 100 Hz dithered random square wave amplitude modulation signal.

## BP 1228405



# market PLACE 

leems mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

## RECHARGEABLE BATTERIES

Although rechargeable nickel cadmium cells are now quite common. it is claimed that the latest range, type NCC, from Ever Ready can be recharged to half their nominal capacity in 60 seconds using standard HP2 cells.

Now available to the constructor, the prices vary according to capacity and quantity ordered. Typically a U2 replacement would cost in the region of $£ 2$ for a single cell but would be proportionately reduced for larger quantities.
Button cells are normally supplied for higher voltage work and typically a $4.8 \mathrm{~V} \quad 550 \mathrm{mAh}$ cell, as used in radio control, would cost approximately $£ 3 \cdot 30$.
Full technical details of the NCC range and stockists can be obtained from Ever Ready Company (Great Britain) Ltd, Special Battery Division. Hockley, Essex.

## INVEST IN CATALOGUES

Time and again we get readers writing in asking where they can obtain certain items, ranging from resistors to infra-red emitting diodes. Apart from telling readers the source of supply we suggest that they should invest in some of the excellent components catalogues available from advertisers.

Typical of these are the latest component catalogues from such firms as: Electrovalue ( 64 pages, 10 p ). Home Radio (Components) Led ( 315 pages, 50 p at shop or 70 p by post), Henry's Radio Lid. ( 350 pages, 40 p at shop. 55 p by post), LST Electronic Components Ltd (44 pages, no charge), and G. W. Smith and Co. (Radio) Ltd. (272 pages, $37 \frac{1}{2} \mathrm{p}$ at shop, $47 \frac{1}{2} \mathrm{p}$ by post).

The catalogues from Home Radio, Henry's and G. W. Smith contain special redeemable vouchers to help off-set the outlay for the catalogue.

The LST catalogue, available free, have added several new items to its opto-electronics section including inexpensive gallium arsenide light sources and planar silicon photo transistors.


Cell construction for the Ever Ready NCC range

There are many other interesting items listed in the LST catalogue including details of their data sheet service.

For our Midland readers, who often complain that they have to order from our London and Southern based advertisers. Hawnt \& Co. Ltd. publish an excellent components catalogue.

Their latest edition (122 pages) contains sections on semiconductors, integrated circuits, capacitors, resistors and ferrites. Further details can be obtained from Hawnt \& Co. Ltd., 112-114 Pritchett Street, Birmingham. B6 4EN.

## VERSATILE PRINTED CIRCUIT BOARDS

The art of being able to pack a large number of circuit components on a small circuit board is a problem that plagues the professional as well as the amateur experimenter.

This problem has certainly been tackled, but whether conquered remains to be seen, by NIP Electronics with the introduction of a range of small printed circuit boards called "Nippiboards".

The board layout is based on the concept that transistor circuits follow common parameters for interconnection, with the addition of occasional requirements for feedback paths and cross-coupling. These paths are provided on the basic interconnection pattern, making it unnecessary to cut away the copper.

Link wires are kept to an absolute minimum and the board layout will, in most cases, follow closely the circuit diagram. This makes for easier circuit checking and servicing.

There are two ranges available: range " $A$ " has an s.r.b.p. base and range " $B$ " which has a fibreglass base. The s.r.b.p. base is cheaper and easy to work with, but the fibreglass base overcomes some of the problems of extreme atmospheric conditions. All boards are flux varnished to keep the copper clean and make soldering easier. Range " $A$ " costs from 15p; range " $B$ " from $18 p$ for one basic pattern.

Typical of the kind of circuits that can be built up on the boards are amplifiers, oscillators, tone circuits, switching circuits, timers, car and home electronic aids. One basic pattern will accommodate up to four transistor stages, whether they be pnp. npn., f.e.t. or u.j.t. devices. Multiple pattern boards are also available for more complex circuits.
To date, the range of boards is only designed for transistor work. But we understand that a range designed specially for integrated circuits will be launched in the near future.

Details of prices and complete range of "Nippiboards" can be obtained from NIP Electronics, P.O. Box 11. St. Albans, Herts.

## Nippiboards from NIP Electronics




The Crab Nebula in the constellation of Taurus. Catalogue No. M1 (NGC 1952)

The remains of a nova observed by the Chinese in 1054 a.d.

It has been established that there are several pulsars and intense X-ray sources in the Crab Nebula. It is expanding at the rate of $1,100 \mathrm{~km}$ per second. The distance from the earth is about 3,500 light years

# ASTRoNOMY IECHIIOUES 

By F. W. Hyde

INN Part 3 the two aerial phase-switched interferometer was briefly described. It is now necessary to deal with this system in more detail since it is the next ste pin the project for those who have the space for it, as the same requirements as to base line apply here. The difference between the simple two-aerial system and the phase-switched system is the extra electronic units necessary and the aerial feeder modifications.

## THE PHASE-SWITCHED INTERFEROMETER

When two aerials are spaced an exact number of wavelengths apart the aerial pattern will be such that the centre lobe will be at a maximum in the plane of symmetry. At this point the pattern is exactly the same as the simple interferometer, where the lengths of the feeders are exactly equal to each other. If now a length of cable equal to one half of a wavelength is introduced in one feeder line the pattern will change so that there is a displacement of the lobes. This displacement will bring the minimum point between lobes to the centre of the plane of symmetry, see Figs. 8.1a, b.


Fig. 8.1. The introduction of one half wavelength piece of cable irito one feeder line displaces the aerial pattern as shown in (b)

As the aerial is moved with the rotation of the earth the aerial pattern can be changed at a predetermined rate so that if there is a source of radiation in the aerial beam it can be scanned.- The alternate "putting in" and "taking out" of the half wave section which does the scanning can be done by using an electronic switching system.

## PHASE SWITCHING SYSTEM

The number of times that the half wave section is switched in and out of circuit is largely a matter of choice, and to some extent depends on the purpose of the project under study. The circuits that will be given will have different speeds of switch operation, but the results on the recorder will not noticeably vary. The record that appears on the chart is an indication of the intensity of the radiation in terms of the modulation of that radiation at the switching frequency. If the switch frequency is, say, 900 Hz , then the radiation received by the aerial will have a component at 900 Hz superimposed upon it. This can be amplified and detected in the system and the result recorded on the charts as shown in Figs. $8.2 \mathrm{a}, \mathrm{b}, \mathrm{c}$. The pattern that appears is the result of the sequence through which the input to the receiver goes before being recorded. This sequence is as follows:

With the source of radiation in the centre of a lobe, that is at maximum sensitivity, the length of the paths will be equal and the aerials will each send to the receiver the same level of signal. The two levels will add together so that the receiver will receive twice the value of the signal. When the half wave section is added the paths now being unequal, the resultant signal will be zero. Therefore the actual amplitude of the signal reverses in phase with the adding and subtracting of the half wave section.

The switching frequency is controlled by a circuit which provides a square wave to operate the switch and at the same time to operate a synchronous detector which reverses the phase of the output in
unison with the switch. The block diagram Fig. 8:3 shows the individual units in which it will be seen that in addition to the units already in operation, there are four new ones.

The switch generator is a multivibrator of the usual type, and may be a part of the unit which includes the narrow band amplifier and the synchronous detector. The actual arrangement is left to the individual constructor.

Each unit will now be described separately, starting with the phase switch for the aerial. There are variations of this: one type being a hybrid junction used as a balanced network made up largely of coaxial cable, and a more simple diode switch unit. Both give similar results but the hybrid junction needs more skill to construct and has to be replaced as a whole when the frequency is changed.


Fig. 8.2. Chart recordings of the sun (a) quiet sun

(b) active sun, showing ionospheric effects at left side of recording

(c) very active sun at the time of a flare, again showing the effects of ionospheric disturbance

## HYBRID JUNCTION PHASE SWITCH

, The hybrid junction phase switch consists of a network of coaxial cable with the switching diodes as part of the assembly, see Fig. 8.4. The action of this switch can be understood by following the diagram and remembering that a "short" looks like an open circuit when seen through a quarter wavelength of transmission line.


Fig. 8.3. Block diagram of the phase-switched interferometer. Units forming the original simple interferometer are shown in unshaded boxes

The switching frequency, which is provided by the switch frequency generator and its amplifier, is fed to the diodes via the choke and capacitor filter network. Suppose that the right-hand diode D2 is conducting, then at the junction of the quarter-wave section with the hybrid it will appear "open", the signal can therefore pass from the input via the first quarter-wave section through the three-quarter wave section to the output point and thence to the receiver.

At this time the left-hand diode D1 is non-conducting and therefore "open"; it appears as a short at the left-hand junction of the hybrid and no signal can pass. When the switch reverses then the left-hand side carries the signal and the right-hand side is blocked off.


Fig. 8.4. A hybrid junction phase switch circuit

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Fig. 8.5. A simple diode switch circuit

## SIMPLE DIODE SWITCH

The simple diode switch consists of a network shown in Fig. 8.5 which uses diodes and a halfwavelength section for the frequency of operation. One advantage of this type of switch is that by having a number of half-wave sections, each for a different frequency, they can be conveniently plugged in to two aerial sockets thus making a very versatile assembly.

The sequence of operation in this switch is that the switching signal is fed into the points 1 and 2 via the filter network: in one direction a diode will conduct, say the left-hand side D1, at this time the other will be non-conducting so that the signal will pass via the conducting arm. When the switching voltage reverses, the diode on the right-hand side will conduct and the signal will pass via the halfwave section. When the switch is in operation the harlf-wave section will be alternately "in" and "out". The same conditions apply therefore as with the more complicated hybrid unit.


Fig. 8.6. Typical circuit for a switch frequency generator

## THE SWITCH FREQUENCY GENERATOR

The switch frequency generator is basically an audio oscillator operating at the frequency chosen for the switch. In this case it is a multivibrator which delivers about 30 volts peak-to-peak with a square wave form, and the circuit is shown in Fig. 8.6. The voltage output is sufficient for the operation of the synchronous detector, but needs a driver amplifier for the phase switch. The reason for this is that the diodes in the phase switch need to pass a current of the order of 50 mA so that the power driver is necessary to accomplish this. The driver amplifier is shown in Fig. 8.7.

## SYNCHRONOUS DETECTOR

As has been explained, the switching in and out of the half-wave section creates a condition where the receiver is fed with a signal which reverses in sign, hence the term phase switch. The reversal is in this case 180 degrees, so a system is required to follow this after the detection and amplification of the switch modulation. The diagram in Fig. 8.8


Fig. 8.9. Typical circuit for a narrow band amplifier set for switch frequency $900-1,000 \mathrm{~Hz}$
shows the details of the circuit for the synchronous detector, and this is fed from the narrow band amplifier and coupled direct to the d.c. amplifier.

## NARROW BAND AMPLIFIER

The narrow band amplifier is an important part of the system because it passes the signal at switch frequency and attenuates those signals which are not required. Strong signals from interference and radio signals which break through from commercial stations are largely eliminated.

The circuit shown in Fig. 8.9 is suitable for the switch frequency adopted. There may be a certain amount of insertion loss due to this amplifier; if this is so and the amplifier has been fed from the second detector of the receiver, then an additional stage of amplification can be interposed between the detector and the narrow band amplifier.

## CONNECTING UP THE SYSTEM

The whole system is now ready to be connected up. It may be that there will be some snags in getting things right straight away. This raises the important point as to the stability of power supplies. If the units are checked individually for operation


A solar radio telescope aerial working at 200 MHz
it should not be too difficult to get the system working satisfactorily. An oscilloscope would be a very useful adjunct in this connection, for all the systems catn then be visually checked. For example the square wave pulses must be of equal width and anmplitude. The phasing of the whole system can be more easily checked when an oscilloscope is used since each successive part of the system caln be dealt with in sequence and any trouble quickly found.

Another point that perhaps should be emphasised again is the importance of good jointing and efficient bonding. Poor earth bonding can lead to many troubles, not least of which is spurious noise and breakthrough. If the area is one where there is a great deal of commercial activity it could be that there will be blocking of the pre-amplifiers. In this case it will be necessary to use some sort of filtering at the aerial input to offset this. Transistor preamplifiers are particularly prone to this trouble.

## ACKNOWLEDGEMENT

The photographs of chart recordings Fig. 8.2 were taken by J. C. Codling, Clare Secondary School.

In the next article some more sophisticated units for the phase-switching system will be described.

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## Gerry Brown winfirnul



## OLFACTRONICS

l've just heard that a Nipponese company are making claim to an invention which amounts to an electronic "nose."

Naturally, smell sensors are not unknown, the Americans developed a device which could smell-out the Vietcong a few years back; this was a kind of funnel-shaped thing containing a microphone which had its fixing adjacent to a thin membrane. The membrane carried a number of sensitive (un-fed) bed-bugs which reacted to the natural body-odours of the enemy in a way which really had to be seen! The "tattoo" which they beat-out on the skin resulted in a signal that could be amplified after the microphone.

The device from Nippon, however, is quite different to previous sensors. For one thing, it can apparently differentiate between a number of different odours, and for another, it is entirely solid-state.

Several compounds are employed in the fabrication of the new device. typica! of which are the oxides of tin, zinc, and the sesquioxide of iron. Whenever such gases as hydrogen, carbon dioxide, acetylene, or propane are applied to the sensor its conductivity decreases sufficiently to give an easily noticed indication, often without further amplification!

We can only hope that the art of olfactronics (my handle) will soon become sufficiently sophisticated to counter the current wave of destruction by detecting the smell of gelignite before it blows someone apart.

## COOL GLOW

Right "out of the blue" (well, (probably nearer purple) someone has come up with a rather interesting way of making a draught. It all came about when Oscar Blomgren of Inter-Probe, Chicago, tried applying an electric field to the task
of diverting an oxy-acetylene flame away from the inside of a blowtorch to prevent its deterioration. Instead, he got a cooling effect!
The cooling results from that funny effect you'll probably remember from experimenting with the school physics lab Wimshurst machine, electric wind. This, as you may recall, has nothing to do with flatulence but is associated with the purply-blue corona discharge that appears around any sharp, or pointed, parts of the machine's electrodes.
The vortex columns of air produced by the discharge are employed by Blomgren to cool the air near any heated surfaces. So efficient is this method of cooling that a glowdischarge produced by a 30 kV potential, at only $200 \mu \mathrm{~A}$, is reported to have already reduced the temperature of an object from 1.675 degrees Fahrenheit to a mere 970 degrees Fahrenheit in a couple of seconds! An ideal device, one would think, where there was a special need for absence of motordriven fans with their attendant vibration. Such noiseless cool must be a real "blow" for fan-makers!


## JABEER-JAMMER

It's quite on the cards that you can tolerate the majority of ambient noises like the intermittent tip-tap of the typewriter 1 am using right now, and the noise from a neighbour's transistor radio. Even so, the chances are that you will experience much more difficulty when attempting to concentrate on some abstruse problem when it is accompanied by the chatter of people in conversation nearby. How to overcome it is a perfectly reasonable question.
Just yesterday, I came to learn the answer to this, seemingly, difficult poser and Fig. I shows the easiest solution. The circuit comprises a simple audio oscillator connected direct to a hearing aid type earphone. The frequency of the device can be varied according to one's particular choice.

This done, it is then only necessary to bring the volume level up to a point which satisfies the compromise between being deafened by the oscillator, and going out of one's tiny mind as a result of the unwelcome intrusion into it by somebody's "yatter".

Fig. 1


## CHILD CARE

The burning need for something that will protect (in particular) the very young from the sometimes malevolent intentions of childsnatchers is a problem that has been in the limelight recently. Even when no harm is intended, the resulting anguish can often, and quite understandably, be unbearable.

From time to time, like others before me, I have made several vain efforts to solve this very real threat to youngsters. Among the ideas have been some which relied upon a weight-sensing switch in the child's pram, or a simple switch arrangement attached to the pram harness. Limitations, unfortunately, exist in these ideas since no facility was provided which sensed movement of the carriage. Perhaps another switch associated with the brake would be the answer; so it goes on, until, ultimately, one is tempted to give the problem up. But, there are quite a few other ways of tackling this enigma.

Not unnaturally, it would be well on the way to meeting our requirements if the abduction alarm had, in addition to warning, ability, a fail-safe feature. It was in this vein that I chanced to come upon an idea that might just be the answer. If the child to be protected carried a low power radio control transmitter (complete with sewn-in antenna) in its coat, a signal would be transmitted, certainly to within the immediate locale of an interested parent.
It could be arranged that "mother" held, at home, a receiver tuned to the signal transmitted from the child and which operated a bell every time the transmitter went out of range. It would then be a simple matter to determine whenever junior was either in the course of being kidnapped, or had just plain toddled off somewhere. Who knows, it might just be better than sending up distress rockets!

Unfortunately, the Post Office is hardly likely to approve of this particular form of "model control"


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|  |  |  |  | 9N7432 | 0.480 .42 | 8 Sa 4140 | 1.801 .76 |
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| L702A |  | $0 \quad{ }^{\text {¢p }}$ | ${ }_{2}^{2 \cdot 60}$ |  | 0.64 0.64 0.60 0.60 |  | 1.40 1.40 1.858 1.858 |
| ${ }_{\text {L } 702 \mathrm{C}} \mathrm{T}$ T05 0.75 |  | $7 \quad 0.72$ | 0.87 | SN7440 | 0.230 .21 | AN74154 | 2.208 .10 |
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$2 \mathrm{~N} 69_{8}$
2 N
$\begin{array}{ll}\text { 2N697 } & \\ \text { 2N698 } & \\ \text { 2N706 } & \\ \text { 2N706A } & 1\end{array}$



## ص\&W Super IC-12 <br> 

## Highfidelity Monolithic Integrated Circuit Amplifier

Two vears ago Sinclair Radionics announced the World's first monolithic integrated circuit $\mathrm{H}_{i}-\mathrm{Fl}$ amplifier, the IC.10. Now we are delighted to be able to introduce its successor, the Super IC. 12 . This 22 transistor unit has all the virtues of the origina! IC. 10 plus the following advantages

1. Higher power.
2. Fewer external components.
3. Lower quiescent consumption
4. Compatible with Project 60 modules.
5. Specially designed built-in heat sink No other heat sink needed.
6. Full output into 3, 4, 5 or 8 ohms.
7. Works on any voltage from 6 to 28 volts without adjustment.
8. NEW 22 transistor circuit.

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Output power 6 watts RMS continuous (12 watts peak).

Frequency Response 5 Hz to $100 \mathrm{KH} \geq \pm$ 1 dB .
Total Harmonic Distortion Less than $1 \%$. (Typical $0.1 \%$ ) at all output powers and all frequencies in the audio band.
Load Impedance 3 to 15 ohms .
Power Gain 90dB (1,000,000,000 times) after feedback

Supply Voltage 6 to 28 volts (Sinclair $\mathrm{PZ}-5$ or $\mathrm{PZ}-6$ power supplies ideal)

Size $22 \times 45 \times 28 \mathrm{~mm}$ including pins and heat sınk.

Input Impedance 250 Kohms nominal.
Quiescent current 8 mA at 28 volts.

With the addition of only a very few external resistors and capacitors the Super IC. 12 makes a complete high fidelity audio amplifier suitawle for use with pick-up. F.M. tuner etc. Alternatively. for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC. 12 ideal for battery operation.


Price, inc. FREE printed crrcuit board for mounting. $£ 2.98_{\text {tree }}^{\text {Post }}$

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Typical Project 60 applications

| System | The Units to use | together with | Cost of Units |
| :---: | :---: | :---: | :---: |
| Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control | £4.48 |
| Mains powered record plaver | Z.30, PZ.5 | Crystal or ceramıc P.U volume control etc. | £9.45 |
| $20+20$ W. stereo amplifier for most needs | $\begin{aligned} & 2 \times Z .30 \mathrm{~s} \text {, Stereo } 60, \\ & \text { PZ.5 } \end{aligned}$ | Crystal, ceramic or mag. P.U., F.M. Tuner, etc. | £23.90 |
| $20+20$ W. stereo amplifier with high performance spkrs. | $\begin{aligned} & 2 \times Z .30 \mathrm{~s} \text {, Stereo } 60, \\ & \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U.F M. Tuner, Tape Deck, etc. | £26.90 |
| $40+40$ W. R.M.S. de-luxe stereo amplifier | $2 \times 7.50 \mathrm{~s}$, Stereo 60 PZ.8, mains trsfrmr | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | MIC, guitar, speakers, etc. controls | £19.43 |

[^2]
# from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules 

## Z. 30 \& Z. 50 power amplifiers



The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02\% at full output and all lower outputs Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well. SPECIFICATIONS ( 2.50 units are interchangeable with $Z .30 \mathrm{~s}$ in allapplications).
Power Outputs
Z. 3015 watts R.M.S. into 8 ohms using 35 volts 20 watts R.M.S. into 3 ohms using 30 volts.
2.5040 watts R.M.S. into 3 ohms using 40 volts 30 watts R.M.S. Into 8 ohms using 50 volts.
Frequency response: 30 to $300.000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$.
Distortion: $0.02 \%$ into 80 hms .
Signal to noise ratio: better than 70 dB unweighted.
Input sensitivity: 250 mV into 100 Kohms .
For speakers from 3 to 15 ohms impedance.
Size: $14 \times 80 \times 57 \mathrm{~mm}$
2.30

Built, tested and guaranteed with circuits and instructions manual. $£ 4.48$
2.50

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tuons manual. £5.48

## Power

## Supply

 UnitsDesigned special for use with the Project 60 system of your choice. Use PZ 5 for normal Z.30 assemblies and PZ. 6 where a stabilised supply is essential
PZ. 530 volts unstabiltsed $£ 4.98$ PZ. 635 volts stabilised $£ 7.98$ PZ. 845 volts stabilised
(less mans transformer) $£ 7.98$
PZ. 8 mains transformer $£ 5.98$

## Project 60 Stereo F.M. Tuner



The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now. Sinclar have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning. printed circuit coils, an I.C. In the specially designed stereo decoder and squelch circuit for silent tuning between stations Good reception is possible in difficult areas, and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.
SPECIFICATIONS-Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz . Capture ratio: 1.5 dB . Sensitivity: $2 \mu \mathrm{~V}$ for 30 dB quieting. $7 \mu \mathrm{~V}$ for lock-in over full deviation. Squelch level: $20 \mu$ V. A.F.C. range: $\pm 200 \mathrm{KHz}$. Signal to noise ratio $:>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ ( $\pm 1 \mathrm{~dB}$ ). Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. Operating voltage: $25-30 \mathrm{VDC}$ Indicators: Power on/tuning/stereo.

Size: $93 \times 40 \times 207 \mathrm{~mm}$.
Built anc tested. Post free.
£25

## Stereo 60 Pre-amp/control unit -00

Designed for Project 60 range but suitable for use with any high quality power amplifier. Again sllicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between charnels. Input selection is by means of push buttons and accurate equalisation is provided fcr all the usual inputs. SPECIFICATIONS-input sensitivities: Radio - up to 3 mV . Mag. p.u. 3 mV : correct to R.I.A A curve $\pm 1 \mathrm{~dB}: 20$ to 25.000 Hz . Ceramic p.u. - up to 3 mV : Aux - up to 3 mV . Output: 250 mV . Signal to noise ratio: better than 70dB. Channel matching: within 1 dB . Tone controls: TREBLE +15 to -15 dB at 10 KHz BASS +15 to -15 dB at 100 Hz
Front panel : brushed aluminium with black knobs and controls. Size: $66 \times 40 \times 207 \mathrm{~mm}$
Butt tested and guaranteed.
$£ 9.98$

## A.F.U. High \& Low Pass Filter Unit



For use between Stereo 60 unit and two 2.30 s or 2.50 s , and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F U. is suitable for use with any other amplifier system. Two filter stages - rumble (high pass) and scratch (low pass). Supply voltage -15 to 35 V . Current - 3 mA . H F. cut-off ( -3 dB ) varrable from 28 KHz to 5 KHz . L.F. cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at 1 KHz ( 35 V . supply ( $0.02 \%$ at rated £5.98 output. Size: $66 \times 40 \times 90 \mathrm{~mm}$. Built testrad and guaranteed.
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## Sinclair Q16/Micromatic

## Q16 High fidelity loudspeaker

The 016 employs the well proven acoustic principles specially developed by Sinclair in which a special driver assembly is meticulously matched to the characteristics of the uniquely designed cabinet. In reviewing this exclusive Sinclair design, technical journals have justly compared the 016 with much more expensive loudspeakers. Its shape enables the Q16 to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures. A solid teak surround with a special all-over cellular foam front is used as much for appearance as its ability to pass all audio frequencies without loss.

This elegantly designed shelf mounting speaker brings genuine high fidelity within reach of every music lover.

## Specifications:

Construction: Special sealed seamless sound or pressure chamber with internal baffle.
Loading: up to 14 watts RMS.
Input Impedance: 8 ohms.
Frequency response: From 60 to 16,000 Hz . confirmed by independently plotted $B$ and K curve.
Driver unit: Special high compliance unit having massive ceramic magnet of 11,000 gauss, aluminium speech coil and special cone suspension for excellent transient response.
Size and styling: 93 in. square on face $x$ $4 \frac{3}{4} \mathrm{in}$. deep with neat pedestal base. Black all over cellular foam front with natural solid teak surround.
Price £8.98.

## Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage AM receiver brilliantly designed to provide remarkable standards of selectivity, power and quality for its size. Powerful AGC counteracts fading from distant stations: bandspread at higher.frequencies makes reception of Radıo 1 easy. The plug-ın magnetic earpiece provided, matches the Micromatic's output to give wonderful standards of reproduction. Everything including the special ferrite rod aerial and batteries is contaned within the minute attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built and tested, you will find it as easy to take with you as your wrist watch, and dependable under the severest listening conditions.

## Specifications:

Size: $36 \times 33 \times 13 \mathrm{~mm}(1.8 \times 1.3 \times 0.5 \mathrm{in}$.) Weight: including batteries, 28.4 gm (1 oz.)
Case: Black plastic with anodised aluminium front panel and spun aluminium dial.
Tuning: medium wave band with bandspread at higher frequencies (550 to $1,600 \mathrm{KHz}$ ).
Earpiece: Magnetic type.
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 or 1；）chm with two inbuilt tweetery and crossover net rork $84 \cdot 20$ ．P．\＆P．30p．E．M．I． $13^{\prime \prime} \mathbf{~ 8 ~}^{\prime \prime}$ twin cone parastatic tweeter）＊ohmit 22.25 ．
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And crontover. 10 watt.
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