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-			
B 38		- 3.2 mm	CHISEL FACE
			_>
B 14	57	— 2.4 mm	CHISEL FACE
D 14	,	- 4.75 mm	SCREWDRIVER
D 24	16	- 4.75 mm	FACE
B 12	2	- 4.75 mm	EYELET BIT
	-		1
B 58	-1-4	- 6.34 mm	CHISEL FACE
LON	162		
	0		
			2
<u> </u>		LIFE	CHISEL FAC
<u> </u>			CHISEL FAC
B 42	LL		D
B 42	LL	1; - 4 75 mm	D
B 42 B 38	LL	³ / ₁₆ − 475 mm	CHISEL FACE
B 42 B 38	LL	+ - 475 mm	CHISEL FACE

Don't take chances. We don't. All our ADCOLA Soldering Instruments are of impeccable quality. You can depend on ADCOLA day after day. That's why they're so popular. You get consistent good service... reliability... from our famous thermally controlled ADCOLA Element and the tough steel construction of this ideal production tool.



* Write for price list and catalogue

PICOLA

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2N5355	Si PNP Hi Gain, low noise, near complement to above	. 3 for 50p
2N2923	25V version of 2N2926. (orange)	7 for 50p
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2N3721	General purpose transistor	8 for 50p

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HI-FI VALUE (1) Garrard SP25 Mk, 111: £11-50, P.P. 50p. (2) Teak Plinth and Tinted Cover: £495, P.P. 35p. (3) Sonatone 9TAHC Diamond Cartridge: £2 50.

1, 2, 3 Bargain Package. £17:95, P.P. 85p.



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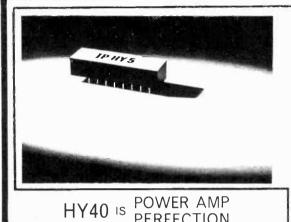
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Electronics) Ltd



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HY40 means Hybrid Power, power neatly locked away inside

A WORLDS FIRST TO JOIN THE WORLDS BEST

The HY5 is a unique and revolutionary concept in High-Fidelity pre-amplifiers. Thanks to the latest techniques, all feedback and equalization networks are, for the first time, combined into an integrated pre-amplifier circuit.

Simply by adding volume, treble, bass potentiometers and only three stabilizing capacitors, which are supplied, your HY5 is complete and ready for use.

The HY5 provides equalization for almost every conceivable input. This years developments in equalization technique enables precise correction for both output voltage and frequency response for any crystal or ceramic cartridge. Yet another feature of the HY5 is its inbuilt stabilization circuit, allowing it to be run off any unregulated power amplifier sunnly.

The HY5 contains a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo preamplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY40 and PSU45 forms a completely integrated system.

INPUTS

Magnetic Pick-up (within ±1db RIAA curve) 2mV Tape Replay (external components to suit head). 4mV Microphone (flat) 10mV. Ceramic Pick-up (equalized and compensatable) 20 - 2000mV variable. Tuner (flat) 250mV Auxiliary 1 250mV. Auxiliary 2 2–20mV OUTPUTS Main Pre-amp output 500mV. 120mV Direct tape output ACTIVE TONE CONTROLS Treble +12db. Bass +12db.

INTERNAL STABILIZATION Enables the HY5 to share an unregulated supply with the Power Amplifier SUPPLY VOLTAGE 15-25 volt SUPPLY CURRENT 5mA approx OVERLOAD CAPABILITY better than 28db on most sensitive

input infinite on tuner and aux! OUTPUT NOISE VOLTAGE 0.5mV PRICE

Mono £3-60 Stereo £7-20





The PSU45 is specifically designed to supply, simultaneously, your HY40 (in mono or stereo format) and one or two HY5s.

Spec.

PSU45 ± 22.5 volts, 2 amps simultaneously.

£4-50 including PRICE: Postage and Packing

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PERFECTION

HY40 is HI-FI POWER ILP are POWER PROUD

only five additional compo-

nents on a printed circuit

board (all of which are

supplied with the HY40).

Power not only for Hi-Fi,

power for Groups, for public

address, for industry, power

for all.

In addition to the P.C. board and manual supplied with the HY40 we now include the five remaining components, at minimal cost, needed to complete the assembly of a High Performance Power Amplifier.

merely combining Βv two HY40s with a Stereo Preamplifier (2 x HY5) and simple Power Supply (PSU45), premium quality stereo may be obtained for a very modest outlay.

The free manual supplied with the HY40 gives clear, easy build instructions for Power Supply; volume, bass, treble and balance controls, together with inputs for Ceramic and Magnetic Pick-ups, and Auxiliary Tape. Tuner functions.

Internally the HY40 is based on conventional and proven circuit techniques developed over recent vears.

1Khz. INPUT SENSITIVITY 300 mV for maximum output. VOLTAGE GAIN 30db at 1KHz.

continuous

FREQUENCY RESPONSE 5Hz-60KHz + 1db TOTAL DISTORTION less than

OUTPUT POWER British Rating 40 WATTS PEAK, 20 watts RMS

LOAD IMPEDANCE 4-16 ohms

INPUT IMPEDANCE 22Kohms at

1% (typical 0.1%) at all output powers. SUPPLY **VOLTAGE** + 22.5

volts D.C. SUPPLY CURRENT 0.8 amps

maximum.

PRICE: including comprehensive manual, P.C. Board and FIVE EXTRA COMPONENTS: MONO £4-40 STEREO £8-80 all post free.

BLACK LIGHT



THE OUT OF SIGHT LIGHT



This kit consists of a high-pressure mercury discharge lamp enclosed in a deep blue-violet outer bulb. This absorbs all visible light given by the discharge but transmits the long-wave ultra-violet rays.

The unit performs in a similar fashion to the fluorescent tube type units in use in some discotheques but instead uses a 3-pin B.C. lamp fitted into a high intensity spot-fitting, with fully adjustable swivel bearing. The reflector is also fully adjustable and the unit may be mounted in any position and focused as desired

White shirts and dresses glow "Whiter than white Paint scenes on walls, etc., with our special fluorescent paint, focus the black-light from across the room and switch-on. The pictures low brilliantly as if by magic!

glow brilliantly as it by magic: Full kit includes lamp, reflector, control-gear, generous samples of five different colours of fluorescent paint and full instructions. Price £25 carr. paid. S.A.E. with all inquiries.

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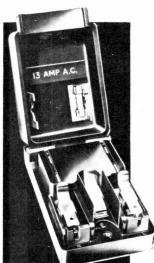
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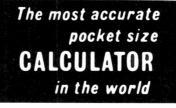
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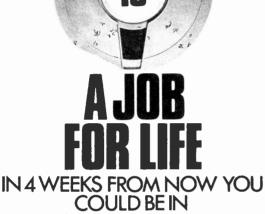
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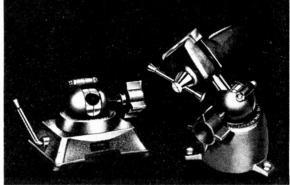
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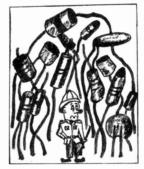
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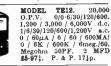
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NEW	''SEW'	' DESI	GNS!	
A A	TYPE SW. 100 100 x 80 mm 500μA \$8:35 1mA \$8:10 20V d.c. \$8:10 50V d.c. \$8:10	TYPE S-80 80 mm square fronts 50μA £3:20 50-0-50μA £3:10 100μA £2:75 100-0-100μA \$3:10	and share and	M 0. 30 30 14 06 47
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can be used to measure a wide range of a.c. and d.c. volt, current and ohms with optional plug-in cards. Specification:--Accuracy: $\pm 0.2, \pm 1$ iligit. Resolution: Inv. No. of digits: 3 plus fourth overrange digit. Overrange: 100% (up to 1-999). Input impedance: 100% (up to 1-999). Input impedance: 100% (up to 1-999). Input impedance: 100% (up to 1-990). Input impedance: 100% (u



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2N3302 17n 2N5310 49n ASV67 45n BE225 19n NKT210 30n ZTX 303 20n	2N3392 17p 2N3310 42p A8Y67 45p BF225 19p NKT210 30p ZTX303 20p 30-30v 24-100 39-100v BA110 25p BY126 12p OA81 8p 2N3393 15p 2N5355 27p ASYK6 82p NKT210 30p ZTX304 25p 24-100 39-100v BA110 25p BN12 12p OA81 8p 2N3393 15p 2N5355 27p ASYK6 82p NKT212 30p ZTX500 15p 25p each 40p each 25p each 8A110 27p BY127 15p OA85 7p OA90 8p 2N402 22p 2N3505 32p ASYK5 30p ZTX500 15p 25p each 40p each 25p each 40p each 25p BA110 27p BA110 27p BA110 27p BA110 27p BA110 27p BA110 B7p BA110 B7p BA110 B7p BA110 B7p
	2N3414 22p 2N5457 30p BC113 10p HFW89 20p NKT217 40p ZTX531 25p 📲 8 A 2 FOR FULL LISTS 📲 BA145 20p BVZ16 40p GA210 17p



Practical Electronics January 1972

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with a 3/16" bit, nickel

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1 amp fuse and booklet



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SK.1 SOLDERING KIT

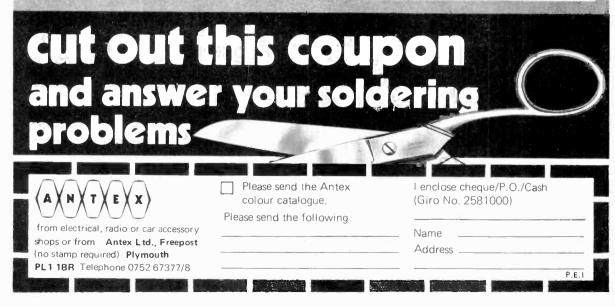
The kit contains a 15 watt 240 volts soldering iron fitted with a 3/16" bit, nickel plated spare bits of 5/32" and 3/32", a reel of solder, heat sink, cleaning pad, stand and booklet "How to Solder". Also available for 220 volts.

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BP04 = 7404 BP05 = 7405	Hex Inverters Hex Inverter (with open-collector	0.12	0.14	0.12
BP05 = 7405				
	output)	0.12	0.14	0.12
BP10 = 7410	Triple 3-input pos. NAND gates	0.12	0.14	0.15
BP13 = 7413	Dual 4-input Schmitt trigger	0-29	0.26	0.24
BP20 = 7420 BP30 = 7430	Dual 4-input pos. NAND gates	0.12	0.14	0.12
BP30 = 7430	8-input pos. NAND gates	0.15	0.14	0.12
BP40 = 7440	output) Triple 3-input pos. NAND gates Dual 4-input Bechnitt trigger Dual 4-input pos. NAND gates B-input pos. NAND gates Dual 4-input pos. NAND buffers BCD to decimal nizie driver	0.15	0.14	0.12
BP41 = 7441	BCD to decimal nixie driver	0.67	0.64	0-58
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	Dual 2-wide 2-input and or-invert	0.15	0.14	0.10
DDIA CHES	Dual 2-sine 2-sine and of invert gates Dual 2-sine 2-sine and of invert gates 2-sine 2-sine and of invert dwide 2-sine and of invert dwide 2-sine and of invert single-phase 3-K flip-flop Dual Master slave 3-K flip-flop Dual Master slave 3-K flip-flop Dual Master slave 3-K flip-flop Dual Atch Dual 3-K with pre-set and clear Gated full adders Dual 2-sine and write memory 2-bit binary full adders BCD decade counter 8-bit shift registers Dual 2-sine 4-bit shift register 4-bit up-down shift register 4-bit up-down shift register 4-bit up-down shift register 4-bit up-down shift register	0.15	0.14	0.12
BP53 = 7453	Quad 2-input expandable and or-			
	invert	0.12	0.14	0.15
BP54 = 7454	4-wide 2-input and or-invert gates	0.15	0.14	0.12
BP60 = 7460	Dual 4-input expander	0.15	0.14	0.12
BP70 = 7470	Single-phase J-K flip-flop	0.29	0.26	0.24
BP72 = 7472	Master slave J-K flip-flop	0.29	0.26	0-24
BP73 = 7473	Dual Master slave J-K flip-flop	0.37	0.35	0.32
$\begin{array}{l} \mathbf{BP72} = 7472 \\ \mathbf{BP73} = 7473 \\ \mathbf{BP74} = 7474 \end{array}$	Dual D type flip-flop	0.37	0-35 0-35	0.32
BP75 = 7475	Quad latch	0.47	0-45	0.42
$\begin{array}{l} \mathbf{BP76} = 7476 \\ \mathbf{BP80} = 7480 \\ \mathbf{BP81} = 7481 \\ \mathbf{BP82} = 7482 \end{array}$	Dual J-K with pre-set and clear	0.43	0.40	0.38
BP80 = 7480	Gated full adders	0.67	0.84	0.58
BP81 - 7481	l6-bit rasil/write wemer	0.07	0.64 0.94	0.88
BP82 - 7489	Phit hinner full addard	0.07	0.94	0.88
BP83 - 7483	Quad full adder	1.10	1.05	0.95
BP82 = 7482 BP83 = 7483 BP86 = 7486 BP90 = 7490 BP91 = 7491 BP92 = 7492 BP93 = 7493 BP94 = 7493 BP95 = 7495 BP96 = 7496	Qual 2 input avaluation NOB action	1.10	1.03	
DP00 - 7400	POD denda sector to NOR gates	0.02	0.30	0-28
PP01 = 7490	BOD decade counter	0.07	0.64 0.84	
BF91 = 7491	8-DIC SHITL registers	0.87	0.84	0.78
BF92 = 7492	Divide-by-tweive counters	0.67	0.64	0.58
BP93 = 7493	4-bit binary counters	0.67	0.64 0.74 0.74	0.58
DP94 = 7494	Dual entry 4-bit shift register	0.77	0.74	0.68
BP95 = 7495	4-bit up-down shift register	0.77	0.74	0.68
BP96 = 7496	ö-bit parallel in parallel out shift-			
	register			0-68
BP100 = 74100 BP104 = 74104	8-bit bistable latches	1.75	1.65	1.55
BP104 = 74104	Single J-K flip-flop equivalent 9000			
	series	0.97	9-94	0.88
BP105 = 74105	Single J-K flip flop equivalent 9001			
	series	0.97	0.94	0.88
BP107 = 74107	Dual Master slave flip-flops	0-40	0-94 0-38 0-53	0.36
BP110 = 74110	Gates master-slave flip-flops	0.55	0-53	0.50
BP111 = 74111	Dual data lock-out flip-flop	1.25	1.15	1.00
BP118 = 74118	Hex set-reset latches	1.00	0.95	
BP119 = 74119	Hex set-reset latches. 24-pin	1.35	1.25	1.10
BP119 = 74119 BP121 = 74121	Monostable multivibrators	0.67	0.64	0.58
BP141 = 74141	BCD-to-decimal decoder/driver	0.67	0-95 1-25 0-64 0-64 1-40 1-70	0.58
BP145 = 74145	BCD-to-decimal decoder/driver O/C	1.50	1.40	1-30
BP150 = 74150 BP151 = 74151	Single J-K hip-nop equivalent successful as the series series and the series of the se	1.80	1.40	1.60
BP151 = 74151	8-bit data selectors (with strobe)	1.00	0.95	0.80
BP153 = 74153	Dual 4-line-to-1-line data	1.20	0.95 1.10	0-95
BP154 = 74154	4- to 16-line decoder	1.80	1.70	1.60
$\begin{array}{l} \mathbf{BP155} = 74155 \\ \mathbf{BP156} = 74156 \\ \mathbf{BP160} = 74160 \end{array}$	Dual 2- to 4-line decoder	1.40	1.30	1-20
BP156 = 74156	Dual 2- to 4-line decoder O/C	1.40	1.30	1.20
BP160 = 74160	Sync, decade counter	1.80	1.30 1.30 1.70	1.60
BP161 = 74161	Sync. 4-bit hinary counter	1.80	1.70	1.60
BP190 = 74191	Sync undown BCD counter	3.50	1.70 3.25	3.00
BP191 = 74191	Sync. binary up-down counter (single	9.90	0.70	0.00
	alook line)	3.50	2.05	3.00
BP192 = 74192	clock line) Sync. up down decade counter	010	3.20	1-75
	Sync. up-down decade counter	2.10	1.80	7.49
DI 100 = 74193	Sync. binary up-down counter (tow clock lines)	0.10	1.05	1.75
BB106 - 7(106	CIOCK IIDES)	2.10	1.95	1.70
Dr 190 = 74190	rre-setable DUMHz decade counter	1.80	1.70	1.60
DF197 = 74197 DD100 = 74200	Fre-setable 50MHz binary counter	1.80	1.70	1.60
Dr 198 = 74198	o-Dit parallel L-R shift register	0.00	5.00	4.00
Dr 199 = 74199	o-bit parallel access shift register	5.50	5.00	4.00
application (TTT	Sync. binary up-lown counter (tow clock linea) Pre-setable 50M Hz decade counter 8-bit parallel L-R shift register 8-bit parallel access shift register mixed to qualify for quantity price. Larg 74 Baries orbit.	er quan	uties—pr	ices on

Devices may be mixed to qualify for quantity price. Larger quantities—prices ou application (TTL 74 Series only). Data is available for the above series of 1.0°s in booklet form. Price 13p. Owing to the ever increasing range of TTL 74 Series, please check with us for supplies of any devices not listed above, as it is probably now in stock.

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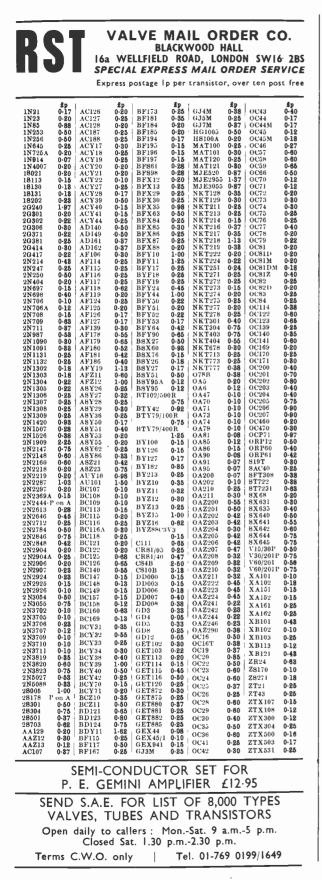
					Price	
Type No.	Case	Leads	Description	1-24	25-99	100 սթ
BP 201C-8L20IC	TO- 5	8	G.P. Amp	68p	53p	45p
BP 701C-SL70IC	TO-5	8	OP Amp	63p	50p	45p
BP 702C-8L702C	TO-5	8	OP Amp Direct OP	63p	50p	45p
BP 702-72702	D.I.L.	14	G.P. OP Amp (Wide			
			Band)	53p	45p	40p
BP 709	D.I.L.	14	High OP Amp	53p	45p	400
BP 709P—µA709C	TO-5	8	High Gain OP Amp	53p	45p	40p
BP 710-72710	D.I.L.	14	Differential			
			comparator	ö3p	45p	40p
BP 711—μA711	TO-5	10	Dual comparator	58p	50p	45p
BP 741 -72741	D.I.L.	14	High Gain OP Amp			
			(Protected)	75p	60p	50p
μA 703CμA703C	TO-5	6	R.FI.F. Amp	43p	350	27p
TAA 263-	TO-72	4	A.F. Amp	70p	60p	55p
TAA 293-	TO-74	10	G.P. Amp	90p	75p	70p
TAA 350	TO-5	8	Wide loud limiting			
		-	amplifier	170p	158p	150p

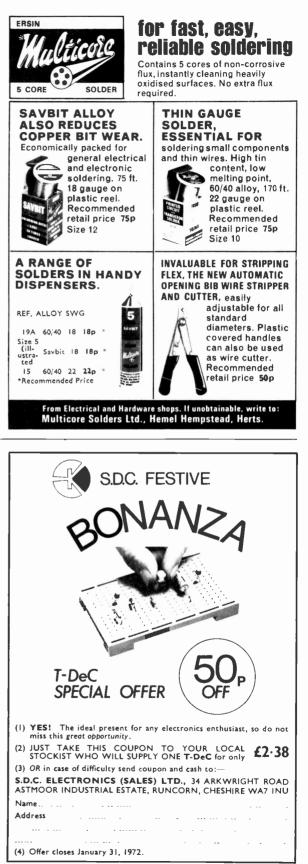














28watts,r.m.s.40Hz to 40kHz ±3dB



RASS

There are two stereo amplifiers-the R100 for ceramic cartridges, the R101 for magnetic and ceramic. Both incorporate FETs (FIELD EFFECT TRANSISTORS), just like top-priced units. FETs give you more of the signal you want, and almost none of the background hiss you don't. Both units have a jack socket to plug in headphones and there's a separate output for tape recorder. Filters (an unusual feature in this price range) and tone controls give a wide range of bass and treble adjustment which compensate for input deficiencies and domestic acoustic conditions.

PRICES SYSTEM 1

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P.U. 011 RADIO WHE UME

Viscount III

Viscount III R101 amplifier	r £22+90p P.&F	λ.
2×Duo Type II speak)
Garrard SP25 Mk. III w		•
MAG. cartridge, plinth a		
cover	£23 + £1·50 P.&F	Ρ.
т	Total £59	
Available complete for		
	P.&F	٩
SYSTEM 2		
Viscount III R101 amplifie	r £22 – 90 n P & F	>
2×Duo Type III speak		•
Garrard SP25 Mk. III v	with	
MAG. cartridge, plinth a	and	
cover	£23 + £1 50 P &F	۰.
cover	£23 + £1 50 P.&F —–	`
	£23+£1·50 P.&F Total £77	,
		,
٢	Total £77	
	Total £77	

SYSTEM 3

Viscount III R100 Amplifier 2 × Duo Type II speakers		
Garrard SP25 Mk. III with CER. diamond cartridge.		,
plinth and cover	£21+£1·5() P.&P.

Total £52

Available complete for only £49+£3-50 P.&P.

SPEAKERS Duo Type II

PFRIE

Size approx. 17 in \times 10³/₄ in \times 6³/₄ in. Drive unit 13in \times 8in with parasitic tweeter. Max. power 10W, 3 ohms. Simulated Teak cabinet. £14 pair + £2 P. & P. Duo Type III Size approx. 231 in × 111 in \times 9½ in. Drive unit 13½ in \times 8¼ in with H.F. speaker. Max. power 20W at 3 ohms. Frequency range 20Hz to 20kHz. Teak veneer cabinet. £32 pair + £3 P. & P.

SPECIFICATION R100/101

SPECIFICATION R100/101 14 watts per channel into 3 to 4 ohms. Total distortion @ 10W @ 1kHz 0.1%. P.U.1 (for ceramic cartridges) 150mV into 3 Meg. P.U.2 (for magnetic cartridges) 4mV @ 1kHz into 47K equalised within ± 1 dB R.I.A.A. Radio 150mV into 220K. (Sensitivities given at full power.) Tane out facilities: beadobase socket power Tape out facilities; headphone socket, power out 250mW per channel. Tone controls and filter characteristics. Bass: +120B to --17dB @ 60Hz. Bass filter: 6dB per octave cut. Treble control: treble +12dB to --12dB @ 15kHz. Treble filter: 12dB per octave. Signal to noise ratio: (all controls at max.) R101--P.U.1. and radio-65dB. P.U.2-58dB. R100 same as R101 but P.U.2 (for crystal cartridges) 450mV into 3 Meg. Cross talk better than ---35dB on all inputs. Overload characteristics better than 26dB on all inputs. Size approx. 13³/₂ × 9in × 3³/₂in.

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	LJ		MULLA	RD,	TEXAS,	FER	RANTI,	GE
l	AC107 AC126 AC127	37p 25p 25p	BSY29 BSY95A BY100	25p 15p 20p	NKT781 NKT10339 NKT10419	29p 25p	IN 4007 IN 41 48	20p 7p
l	AC128 AC151 AC176	20p 12p	BYXI0 BYZI0	15p 40p	NKT10439	19p 27p 22p	2G302 2G371	19p 15p 25p
۱	AC176 AC187	25p 30p	BYZI2 BYZI3	30p 20p	NKT20329 0013	31p	2G374 2N174 2N385A/	80p
١	AC187 AC188 ACY17	30p 29p		150	NKT80111 NKT80112	67р 83р	2N385A/ 2N388A 2N404	75p 23p
ł	ACY18 ACY19	20p 20p	C3V3 C3V6 C3V9 C4V3 C4V7	15p 15p	NKT80113 NKT80211 NKT80212	75p	2N696 2N697	15p
l	ACY19 ACY20 ACY21 ACY22 ACY40 ACY41	19p 19p 19p	C4V3 C4V7	15p 15p 15p	NK 180212 NK 180213 OAS	75p 75p 20p	2N698 2N706	30p 10p
	ACY40 ACY41	150	C5V1 C5V6 C6V2	15p 15p	0A10 0A47	25p 8p	2N706A 2N708	12p 16p
1	ADI40	55p 57p	C6V8 C7V5 C8V2	15p 15p	OA70 OA73	8p 8p	2N711 2N711A	37p 37p
1	AD161 AD162	37p 37p	C9V1	15p 15p	0A79 0A81	8р 8р	2N911 2N914 2N918	50p 20p 42p
I	AF106 AF114 AF115	24p 25p 25p	C10 C11	15p 15p	O A 85 O A 90	8p 8p	2N1090 2N1091	30p 33p
	AFII6 AFII7	25p 25p	CI3	15p 15p 15p	OA91 OA95 OA200	8p 8p 10p	2N1131 2N1132	30p 30p
	AF118 AF124 AF126	44p 25p	C16 C18	15p 15p	OA202	10p 37p	2N1302 2N1303	20p 20p
ľ	AFI39	17p 33p	C10 C11 C12 C13 C15 C16 C16 C20 C22 C24 C27 C30 C15	15p	OC20 OC22 OC23 OC24	97р 47р	2N1304 2N1305	25p 25p
	AF186 AF239	40p 37p 25p	C24 C27	15p 15p	OC23 OC24	60p 60p	2N1306 2N1307	30p 30p
Ē	ASY26 ASY27 ASY28	30p 22p	01311	15p 45p	OC25 OC26	37p 33p	2N I 308 2N I 309 2N I 507	34p 31p
	ASY29 ASZ21 AUY10	30p 37p	MJE520 MJ480 MJ481	75p 97p ≰1-25	OC29 OC35	60p 75p 50p	2N1507 2N1613 2N1711	23p 22p 25p
	AUYI0 BAI15	£1.50 8p	MJ490 MJ491	£1-00 £1-35	OC24 OC25 OC26 OC28 OC29 OC35 OC36 OC41 OC42 OC44	63p 25p	2N2147	82p
	BC107 BC108	2p 2p 2p	MPF102 MPF103	43p 37p	OC42 OC44	30p 15p	2N2148 2N2160 2N2368	63р 62р 17р
	BC109 BC147 BC148	150	MPF104 MPF105	37р 40р	OC45 OC71	15p 15p	2N2369 2N2369A	17p 20p
	BCI49 BCI58	15p 17p	NKTI24 NKTI25	30p 40p	OC72 OC75	23p 23p	2N2646 2N2904	50p
	BA115 BC107 BC108 BC109 BC147 BC148 BC149 BC158 BC167 BC168 BC169 BC169C BC182 BC182L	11p 10p	NKT126 NKT128 NKT135	37 p 25 p 26 p	0C44 0C71 0C72 0C75 0C76 0C77 0C81 0C81D 0C817	25p 40p 23p	2N2904A 2N2905	49p 65p
	BC169 BC169C	p 9p	NKT137 NKT210	32p 25p	OC81D OC81Z	20p 55p	2N2905A 2N2906	75p 44p
l	BCI82L	12p 10p 9p	NKT211 NKT212	25p 25p	OC81Z OC82 OC82D	25p 15p	2N2906A 2N2926 al colours	54p
l	BC182L BC183 BC183L BC184L BC2124 BC212L BCY30 BCY31 BCY32 BCY33	9p 15n	NKT213 NKT214	25p 23p	OC82D OC83 OC84 OC139 OC140 OC170 OC171	23p 25p	2N 3053 2N 3054	25p 63p
	BC184L BC212	15p 17p	NKT215 NKT216 NKT217	21p 46p 50p	OC139 OC140	25p 35p 25p	2N3055 2N3702	75p 11p
	BC212L BCY30	12p 25p	NKT218 NKT219	25p 25p	OC171 OC200	30p 37p	2N3703 2N3704	10p
	BCY32 BCY33	48p 50p 20p	NKT223 NKT224	27p 25p	OC200 OC201 OC202 OC203 OC204 OC205 OC206	47p 63p	2N3705 2N3706	10p 9p
	BCY33 BCY34 BCY38	25p 30p	NKT225 NKT229	21p 29p	OC203 OC204	37p 40p	2N3707 2N3708 2N3709	Пр 7р 9р
١	BCY70 BCY71 BCY72	19p 37p	NKT237 NKT238 NKT239	31p 19p 23p	OC205 OC206	65p 75p	2N3710 2N3711	9p 9p
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	BF195 BF196 BF200	15p 35p	NKT351 NKT401 NKT402	71p 77p	ZTX320 ZTX330	30p 18p	40309	33p 45p
	BFX13 BFX29	25p 31p	NKT403 NKT404 NKT405	65p 60p 79p	Z1X500 Z1X501	16p 16p	40312 40320	48p 36p
ł	BFX84 BFX85	26p 34p	NKT405 NKT406 NKT420	62p €I-83	ZTX 502 ZTX 503	17p	40360 40361 40362	43p 48p 58p
	BFX86 BFX87 BFX88	25p 30p	NKT451 NKT452	58p 54p	ZTX 504 IN34A IN60	40p 20p 20p	40406	56p 39p
	BFY50 BFY51	25p 23p 19p	NKT453 NKT603F	50p 30p	IN64 IN82A	20p 47p	40408 40409	51p 54p
	BFY52 BFY53	20p 16p	NKT613F NKT674F	30p 30p	IN914 IN4001	7р 7р	40410 40468A	62p 35p
	BFY90 BSX19	67p 16p	NKT676F	30p 28p 29p	IN 4002 IN 4003	100	40600 40601	58p 55p
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7413 Schmitt Trigger 351 35 7420 Dual 4-input NAND		triggered flip-floj Quad bistable latch	p 141 -	46р 38р 45р 40р	33p 37p		counter with on line mode contro	e ⊳i20i ∡i-80	£1-48	£1-27
gate 2 20 7430 8-input NAND gate 3 20	o 16p I4p 7476	Dual J-K master-		.sp isp	57 p	74191	Synchronous up down 4-bit binar	y		
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TRADE WINDS

 \mathbf{F} OR some time it has been clear that certain sections of the U.K. electronics industry will face a difficult. if not threatening, situation due to overseas competitors, chiefly Japan, in the coming years. For the home producers of television receivers, it seems the danger is quite imminent.

Shut out of the American market by recently imposed tariffs and debarred from Germany and Italy by the terms of the licensing agreement which allows them to manufacture television receivers for the PAL system. the Japanese look now to the large U.K. market as a main outlet for their huge output of colour receivers. And with 20in models likely to retail at £100 against home products costing around £180, it is too much to expect patriotism to win against price. Anyhow pre-cedents have already been provided by the watch and camera trades, and imported radio, audio, and measuring equipments are now taking over sections of the home electronics market.

There is, unfortunately, the possibility of an even more serious threat to the electronics industry in the next decade. The danger of an unimpeded flow of cheap products is not merely the worry of that part of the industry catering for the consumer market. Producers of industrial equipment may well have to face intense overseas competition in the coming years. Japan, for example, is reported to be making phenomenal progress in industrial electronics. She has already taken over the U.S.A. market for desk calculators; and as the largest ship building country in the world, she can be expected to exploit this situation fully, and to offer in time vessels fully equipped with communications, radar, and navigational equipment, as complete going concerns. This kind of development in world trade could be a serious blow to the capital equipment section of the U.K. electronics industry-the very section which has so far been the major exporter, in terms of value, of the whole U.K. electronics industry.

Can any measures be taken to avert the more serious consequences of this "cut price" competition from countries operating under quite different social and economic systems to our own? British industrialists no doubt look with envy to the U.S. and would welcome a similar initiative from our government. But tariffs and trade protection arouse no enthusiasm among customers at large. More likely our industry's chief hope rests in the continuing expansion of world requirements for electronic products of all kinds. But frequent "priming of the pump" by the introduction of technical innovations that lead to further new and commercially valuable applications for electronic techniques seems now, more than ever, an imperative operation for the industry. F.E.B.

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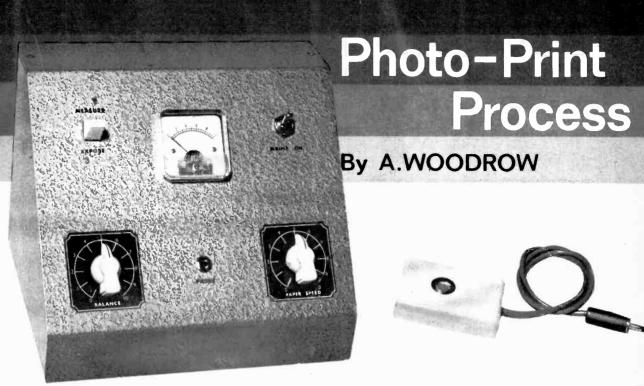
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Our February issue will be published on Friday, January 14.

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MANY amateur photographers like to do their own processing, and an increasing number are attempting colour printing. One of the more tedious tasks of processing is the printing of a roll of negatives. When printing monochrome, each negative requires that a test strip be made to determine the correct exposure needed, the test then being assessed and a final print then made.

When a roll of 36 negatives is being printed, the amount of time and wasted materials involved can be quite considerable. When colour negatives are used, the problem is multiplied by the necessity of obtaining the correct colour balance as well as density. It is possible to waste a complete printing session without obtaining a single satisfactory colour print.

The unit to be described is designed to take most of the trial-and-error out of printing; in fact, prints are made almost automatically in the case of monochrome. The work involved in colour printing is a little greater than for black and white, but it is still possible to make a successful print at the first attempt.

THREE FUNCTIONS

An exposure meter first assesses the negative, and the information obtained is transferred to a timer, which switches the enlarger lamp on for the required time to give a correctly exposed print. The exposure meter is also used for the comparison of a colour negative with a "standard" negative to obtain the correct colour balance.

One of the most important factors in successful photographic processing is consistency of conditions; temperatures, strength of solution, etc. must be constant if good results are to be obtained. One such factor which is often overlooked is mains voltage. If an enlarger lamp is operated at low mains, its light will be more yellow than at nominal voltage, as well as giving a reduced light output. Conversely, a lamp operated at high voltage gives a bluer light.

The change in light output is serious enough in monochrome printing, but the additional variation in colour can cause chaos in colour printing; a change of two or three volts is sufficient to cause a noticeable shift in colour. With this in mind, the control unit includes a stabiliser unit to hold the lamp voltage constant against variations of mains.

The control unit thus has three parts, intended to be used together, although any one or two sections could be constructed and used independently of the others. The unit will automatically assess and expose negatives over the range one second to one minute, and the stabiliser holds the lamp voltage constant for mains inputs between 220–250V. For nominal mains voltages other than 240V, the stabiliser will operate over a correspondingly lower range.

EXPOSURE METER

If a light source of known brightness is projected through a negative, the amount of light allowed to pass by the negative is a measure of the density of it, and thus of the exposure required to make a print. In this case, the light source is the enlarger lamp, and the amount of light passed by the negative is measured at the enlarger baseboard by a light dependent resistor.

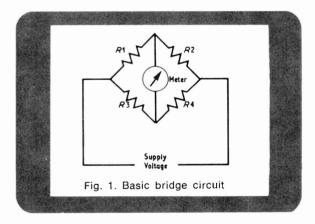
This device, as its name implies, varies its resistance according to the amount of light falling on its sensitive surface. When the device is placed on the baseboard, its resistance depends on the amount of light transmitted by the negative, and thus on the required exposure.

A light dependent resistor, however, has no "memory"; when the illumination is removed, the information gained from the measurement is lost. Control Unit

As this information is later needed by the timer, a further element must be added to store the reading, ready for transfer to the timer. A bridge is therefore used for the measurement, and the basic circuit is snown in Fig. 1.

BRIDGE CIRCUIT

The main property of this circuit is that when $R_1 \times R_4 = R_2 \times R_3$ no current will pass through the meter. In this condition, the bridge is balanced. When the equation is not satisfied, a current flows in the meter. This current is of a magnitude dependent on the degree of unbalance and the voltage of the supply, but the point at which balance occurs is independent of applied voltage.



Referring now to the main circuit in Fig. 2, the bridge is formed by VR1, VR2 and the light dependent resistor PCC1. Resistors R10, R11 and R15 are also included in the bridge, but serve only to limit the range of the controls, preventing any arm of the bridge from being short-circuited. Potentiometer VR1 forms two arms, either side of its wiper, this control being preset according to the type of printing paper in use.

The supply is 12V a.c., taken from the mains transformer. The output from the bridge is rectified by D3, and then passed to TR4. This transistor is connected as an emitter follower, i.e. a current amplifier, driving ME1, a ImA meter. Resistor R14 limits the maximum current through the transistor, preventing the meter needle from wrapping itself around the endstop if the unit is switched on with the bridge well out of balance. C3 removes ripple, giving a steady meter reading, and R12 adds a small amount of forward bias to TR4. The function switch S2 selects either "measure" or "expose" operations. When negative measurement is required, S2c and S2d switch the balance potentiometer into the bridge circuit. Closure of S2b applies voltage to the meter circuit, and also switches on the relay driver transistor TR3 via R8, thus energising the relay RLA and turning on the enlarger lamp for focusing and measurement purposes. The diode D2 prevents voltage surges from damaging TR3 when the relay is de-energised.

TIMER

With the function switch in the "expose" position, S2c and S2d transfer the information regarding the required exposure, contained in VR2, into the timer circuit. As S2b is now open, the relay is de-energised. No supply voltage is yet applied to the timer, as RLA1 and S2e contacts are open-circuit at this stage.

Contacts S2e are now closed to initiate the timing period. Resistors R2, R3 and R4 form a voltage divider, and TR2 has a forward bias derived from the voltage developed across R4. This transistor is thus turned on, its collector volts now being close to the supply of 12V.

The voltage dropped across R5 is applied via R6 to the base of TR3, energising the relay. The enlarger lamp is thus turned on via contacts RLA2, and contacts RLA1 bypass S2e which may now be released, as the supply voltage to the timer is now applied by the relay contacts.

THE NEED FOR AN F.E.T.

The heart of the timer is C1, VR2 and TR1. a field effect transistor. The main property of a field effect transistor in this application is its extremely high input resistance. The highest value of timing resistance which can be used in any timer circuit is determined by the resistance of any circuit connected to it.

If a bipolar transistor were used for TR1, the maximum input resistance obtainable would be only a few megohms. For the timing period to be determined by the timing resistance alone, as is required for accurate peformance, the maximum value of timing resistance must be about one tenth of the input resistance of TR1, i.e. a few hundred kilohms.

As the timing period is proportional to $C1 \times VR2$, and the timing resistance is low, C1 must obviously be high. Large values of capacitance give rise to further problems, as they will have a relatively low leakage resistance, and this can also affect the accuracy of the timer. The field effect transistor, with an input resistance of several hundred megohms, allows a low value to be used for C1, with a large value of VR2.

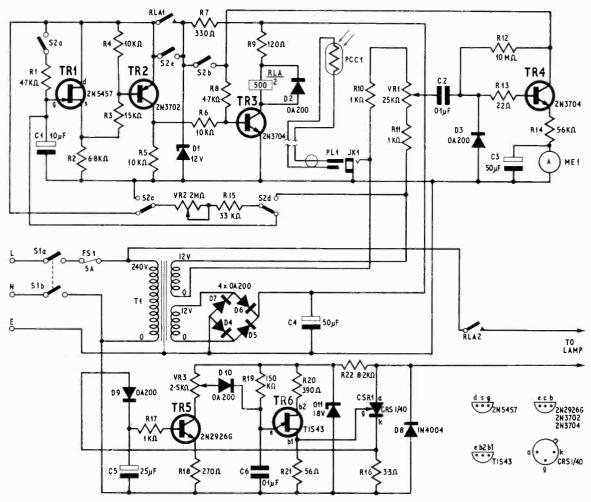


Fig. 2. Complete circuit diagram of the enlarger control unit

THE TIMING OPERATION

Returning to the operation of the complete timer, TR1 has a small bias applied by the voltage drop across R2. Initially C1 is discharged, so the gate of TR1 is at zero volts. When the supply is connected by S2e. C1 begins to charge at a rate determined by the setting of VR2, previously determined by the exposure meter. As the gate voltage approaches the bias voltage of the source, TR1 begins to conduct. The gate voltage thus becomes transferred to the source by the self-biasing action of the source current. Resistor R3 also applies the source voltage to the base of TR2.

When TR1 source voltage, and thus TR2 base voltage, reaches approximately 11V, TR2 switches off, this transistor being a *pnp* type. The collector voltage falls to zero, switching off TR3, and deenergising RLA. The enlarger lamp is turned off, and the supply is removed from the timer by contacts RLA1, S2e having been previously released.

The standing bias, derived from the potential divider R2-R3-R4, is removed from TR1, which now becomes forward biased by the charge remaining on C1. The capacitor is discharged rapidly through the gate-source junction of TR1 and R2, resetting the timer ready for the next timing operation.

THE FUNCTION OF R1

The function of R1 is best explained by considering the circuit without this component. During the measurement operation, VR2 is disconnected from the gate of TR1, leaving C1 in a discharged condition, but voltage is applied to the timer via the contacts of RLA1 which is energised.

When the function switch is changed to "expose", the timer begins a complete timing sequence, as C1 is now charged via the nowconnected VR2. The enlarger lamp will thus remain on until a complete timing sequence has elapsed, which is obviously inconvenient if the timer is set for more than a few seconds.

During measurement, R1 is switched in, maintaining C1 in a fully charged condition. When the function switch is changed to the "expose" position, the timer is in the same state as at the end of a timing period, and the lamp is switched off.

The power supply for the exposure meter and timer is conventional. A full-wave bridge rectifier. D4 to D7, is fed from the mains transformer T1, the resulting pulsed d.c. being smoothed by C4. The relay is then supplied via R9, and the remainder of the circuits are fed from a 12V rail stabilised by R7 and the Zener diode D1.

LAMP STABILISER

The basic principle of the stabilised supply to the enlarger lamp is illustrated in Fig. 3a, which shows two cycles of the voltage across the lamp. Only part of each positive half-cycle is applied to the bulb, although the whole of each negative half-cycle is applied.

If the supply is fed to the lamp at point A, the result will be a higher r.m.s. voltage for the complete cycle than if the supply were not allowed to reach the lamp until point B.

Thus, a rise in mains voltage can be counteracted by delaying the point in the positive half-cycle at which the bulb is switched on. Conversely, a drop in mains voltage can be corrected by an earlier switch-on, allowing more of the half-cycle to reach the bulb.

Referring to Fig. 2, each negative half-cycle is fed to the lamp via D8. The positive half-cycles are controlled by the silicon controlled rectifier CSR1.

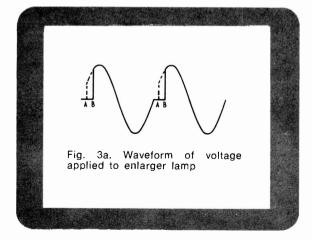
The silicon controlled rectifier, or thyristor, is an extension of the normal silicon diode. It will under no circumstances conduct when the cathode is positive with respect to the anode, and will also block when forward biased, until a third electrode, the gate, is made positive with respect to the cathode. The thyristor will then conduct until the forward voltage is reduced to zero, even though the gate voltage is reduced.

Once the thyristor turns off, it will revert to its blocking state until it is again forward biased and a gate voltage is applied. Thus CSR1 can be switched on at any point during the positive half-cycle as desired, by applying a gate voltage when required. The gate voltage is derived from TR6, a unijunction transistor.

THE UNIJUNCTION TRANSISTOR

The unijunction is a semiconductor device with two bases and an emitter. It acts as a pure resistance of 5 to 10 kilohms with zero emitter voltage. This condition is maintained as the emitter voltage is increased, with only a few microamps of emitter current, until a critical (peak point) level is reached.

At this point, the emitter-base one junction breaks down, allowing a large emitter current to flow. If the emitter voltage is now removed, the device reverts to its original state, with very small emitter current, and once more acts as a resistance between base one and base two.



TRIGGER CIRCUIT

Referring again to Fig. 2, the diode D10, TR5 and the associated circuitry will initially be ignored.

At the beginning of the positive half-cycle, CSR1 is not conducting. D8 is reverse biased, and is also blocking. Thus, the full mains voltage is developed across CSR1, and is applied via R22 to the trigger circuit formed by TR6.

Initially, D11 does not conduct, until the mains voltage reaches 18V, when the Zener will clamp at this voltage. Thereafter, the supply to the trigger circuit will remain at 18V, the surplus voltage being dropped across R22.

As soon as the supply begins to rise, C6 starts to charge via R19. As the supply voltage is stabilised by D11, the capacitor will always charge at the same rate. When the charge on C6 reaches the peak point voltage of the unijunction TR6, the emitter conducts, and C6 discharges through the emitter-base one junction. This current gives a positive voltage across R21, and this voltage is applied to the gate of CSR1, switching it on.

Once the trigger circuit has supplied the gate pulse, it is no longer required until the next positive half-cycle begins. As CSR1 is in parallel with the supply to the trigger circuit, and as this component has a very small voltage drop when conducting, it is apparent that the trigger circuit will be inactive once it has turned on the thyristor, and will remain so during the negative half-cycle, when D8 is conducting.

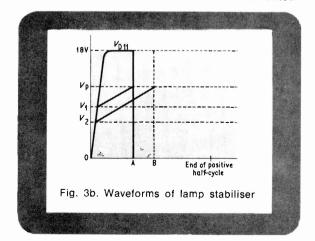
Resistor R20 is included to stabilise the working conditions of TR6 against temperature variations,

CORRECTION OF FIRING POINT

Transistor TR5 and the associated circuitry gives the correction of thyristor firing point necessary for stabilisation against mains variations. The mains voltage is monitored by R16, in series with the lamp and CSR1.

This resistor develops a voltage across it while CSR1 is conducting, i.e. during the positive halfcycle. As this half-cycle is incomplete, being chopped to provide the correction in the lamp supply, it is obvious that the r.m.s. and average voltages across R16 will not provide a suitable reference. The peak voltage is thus used.

The waveform across R16 is applied via D9 to C5. This capacitor is charged to the peak of the waveform, and will remain at that level unless



allowed to discharge. Three discharge paths are available. C5 cannot discharge back into the supply as it is prevented from doing so by D9.

The second path is into the base of TR5. By the use of a high gain transistor in this position, a relatively high input resistance is achieved, and C5 loses little of its charge this way.

The third possible path is through its own leakage resistance, but the loss this way is also small enough to be ignored. Thus, the voltage across C5 is proportional to the peak voltage of the mains supply. This voltage is amplified by TR5, and a portion of the amplified voltage is tapped off from VR3, being passed by D10 to give a standing charge on C6.

STABILISER OPERATION

Taking the action of the stabiliser unit in its entirety, and referring to Fig. 3b, a voltage proportional to the peak of the mains supply appears across C5. This voltage is amplified by TR5, and tapped off from the slider of VR3, then applied to C6. This gives a voltage on C6 of V_1 . C6 is also charged via R19, and its voltage increases until it reaches the peak point voltage of TR6, V_p . The thyristor is then turned on by the pulse across R21, and the remainder of the half-cycle is applied to the enlarger lamp.

If the mains voltage now rises, a larger voltage is developed by C5. This results in a reduced voltage at the collector of TR5, and a lower initial voltage on $C6-V_2$ in Fig. 2b. Then C6 begins to charge via R19 at the same rate as before; as the starting voltage is lower than previously, C6 takes longer to achieve V_p . The thyristor is thus turned on later in the half-cycle, compensating for the rise of mains voltage.

The peak point voltage of unijunctions varies from sample to sample, and an adjustment is necessary to compensate for this. VR3 is therefore used to adjust the d.c. level fed to C6, this being the only control necessary to set up the stabiliser.

ALTERNATIVE SEMICONDUCTORS

Many of the components are uncritical, and alternatives can be used. This applies particularly to the semiconductors.

TR1 must be an *n*-channel field effect transistor. The 2N5457 (also available under the number MPF103) has been selected because it has a lower range of characteristics compared with other types. Other f.e.t.s will work satisfactorily, but may not give the same range of times—this is not too important, as the only result will be a different setting of the paper speed control when the unit is calibrated.

TR2 is required to be a *pnp* silicon transistor, and any such component will work satisfactorily. Similarly, TR3 and TR4 are merely required to be *npn* silicon types, the only limitation being that TR3 must be able to handle the relay current—100mA at most. TR5 is a high gain *npn* silicon transistor— BC109 or 2N3711 are two suitable alternatives. TR6 is a unijunction—2N2646, 2N2160 or BEN3000 will work quite happily here. If alternatives are employed, the lead-outs must be checked, as the various types may differ from those specified.

Any Zener diodes of the required ratings are suitable, and D8 may be any silicon diode with a 1A 400V rating, or better. Similarly, CSR1 can be

COMPONENTS . . .

$\begin{array}{llllllllllllllllllllllllllllllllllll$
All resistors \pm 10%, $rac{1}{4}$ W unless otherwise specified
Potentiometers VR1 $25k \Omega$ lin VR2 $2M \Omega$ log VR3 $2.5k \Omega$ subminiature skeleton preset
Capacitors C1 10 μ F 15V tantalum C2 0.1 μ F 250V foil C3 50 μ F 6V electrolytic C4 50 μ F 25V electrolytic C5 25 μ F 6V electrolytic C6 0.1 μ F 250V foil
Semiconductors TR1 2N5457 (or MPF103) TR2 2N3702 TR3 2N3704 TR4 2N3704 TR5 2N2696G TR6 TIS43 CSR1 CRS1/40
Diodes D1 12V 400mW Zener D2-7 OA200 (6 off) D8 1N4004 D9, 10 OA200 (2 off) D11 18V 1W Zener
Switches S1 2-pole on-off, toggle S2 P.O. type lever switch 4CL/4CN (see text)
Relay RLA 500 ohm coil, 2 make contacts (one heavy duty)
Transformer T1 Miniature mains transformer, 2 × 12V secondaries (R.S. Components)
Miscellaneous PCC1 ORP12 ME1 1mA meter, type MR38P Veroboard, 5in × 3≵in, 0.15in pitch, copper clad Case, 8in × 6in × 6in sloping front FS1 Fuse 5A and fuseholder 2 pointer knobs 3.5mm jack plug and socket 4 rubber feet Heat sink, TO5 Small plastic box Audio screened lead

any type of adequate rating, 1A 400V. The rest of the silicon diodes, specified as OA200, may be any small signal types, such as OA202, 1N914, 1N916, SD19.

RELAY AND FUNCTION SWITCH

The specification of the relay and function switch gives the minimum requirements of these components. It may prove difficult to obtain exactly the contacts required.

The relay may have a different coil resistance from the 500 ohm type specified. Any resistance down to 200 ohms can be used in the circuit as it stands: relays down to 150 ohms may be accommodated by reducing R9 to 82 ohms. Two make contacts are needed; most relays will have two or more sets of changeovers, but the unwanted contacts can, of course, be ignored. One set of the contacts at least must be capable of handling 1A at 240V.

The function switch is a P.O. type lever keyswitch. This has three positions; when the lever is pushed in one direction from the centre, the switch locks in this position; this section requires two changeovers and two make contacts. When the lever is pushed in the other direction from centre, the switch is spring-loaded, i.e. it will return to centre when released; this section requires one make contact. As with the relay, any spare contacts can be ignored.

A suitable switch, with four changeovers in each direction, is available from several suppliers, for example Home Radio. The older type of lever keyswitch, with a cylindrical knob is also suitable. As a last resort, the function switch may be separated into two parts; a four changeover wafer switch is used for the "measure" part, and a separate pushbutton of the press-to-make, release-to-break type initiates the timer.

OTHER COMPONENTS

Alternatives to the specified components must not be used in two positions. C1 must be a low leakage capacitor, tantalum being the most suitable type. Paper capacitors are available, but tend to be bulky and expensive. The more familiar aluminium electrolytic should not be used; in fact, the timer will fail to operate with such a component.

This unit requires a transformer with two separate secondaries. Most of the small mains transformers available have a single centre-tapped winding, i.e. 12V-0-12V, but these should not be used. The transformer specified has the necessary secondaries.

The value of R16 must be changed according to the size of the enlarger lamp. A 150W bulb requires 3·3 ohms in this position; 6·8 ohms is used with a 75W lamp.

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IN ELECTRONIC ENGINEERING

By A. Freeman, B.Sc., A.R.I.C. (The Marconi Company Ltd)

A LTHOUGH, during- the last two decades, the development of transistors and printed circuitry has led to considerable economy of space per equivalent circuit function, the requirement for more sophisticated apparatus has meant that electronic capital equipment now in use may occupy about the same volume as did the simpler systems of 20 years ago. Thus, the reduction in the consumption of metal by the electronic capital goods industry, that would have been expected as a result of the economy of space, has not taken place.

There has, however, been an increase in the use of aluminium alloy because of its lightness, workability, and better corrosion resistance than steel. For the chassis or cases in which circuits are built, cadmium plated steel is often used because of the difficulty of making solder joints in aluminium.

RADIO TELESCOPES

Radio astronomy and communication satellites have led to the erection of large steerable reflector "dishes." Such "dishes" are made from aluminium alloy or stainless steel and must be supported so that their profile is accurately maintained, even in galeforce winds.

Cable and Wireless satellite-communication earth terminal at Bahrain with 90ft diameter dish antenna Thus, there has grown up a heavy constructional industry, ancillary to the electronics side of the business, using the techniques and materials of ship and bridge-building, but with the accuracy of light engineering.

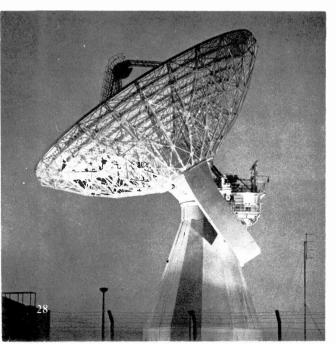
When these devices are erected in climates where winter temperatures are appreciably below freezing point, girders made from ordinary mild steel cannot be used, because at these temperatures they become brittle; comparatively light mechanical shocks can result in fracture. Therefore special steels are used which retain their toughness in very cold conditions.

PRINTED CIRCUITS

Copper is still the most used electrical conductor, not only in the form of wire, but in increasing quantities on printed circuits, where the connections are made by etched foil on an insulating baseboard.

The copper foil for the manufacture of the clad laminates is made by electro-plating polished stainless steel strip and parting the two metals mechanically, which is not difficult because the electro-deposit has only poor adhesion to the polished surface. This method produces foil that is polished on one side and dull on the other.

Erection of the aerial for satellite communications on Ascension Island





The dull side is coated with resin and applied to the surface of the insulating laminate at the same time as fabricating the laminate. The roughness of the dull surface of the copper foil acts as an adhesion key.

Connections are made to the printed circuit either by soldering or, more generally, by sockets containing spring contacts which mate with circuit terminations on the edge of printed-circuit boards. The contacts and the circuit terminations are often goldplated to prevent tarnishing, which would otherwise lead to poor contact and "noise" in the circuit.

THIN-FILM CIRCUITS

Thin-film devices are similar to printed circuits in that all interconnections are laid down, but they also contain resistors and capacitors as integral parts of the circuit. These devices are made by evaporation techniques. The substrates are glass or ceramic; the metal to be used is evaporated in a vacuum with the metallic vapours passing through a mask to condense in the required pattern on the substrate. The substrate must be kept perfectly clean.

The materials used in this process include gold for the circuit nichrome for resistors, and silicon monoxide as insulation. Capacitors are built up by successive layers of metal and silicon monoxide. As in the case of printed circuits, the manufacture of thin-film devices readily lends itself to automation.

WAVEGUIDES

Brass is used in the manufacture of waveguides, particularly those of large aperture, which are made from sheet as opposed to the smaller drawn tube. In areas where ferrous metals cannot be used because of electro-magnetic fields, brass screws are employed and springs are made in either phosphor-bronze or, for more severe duty, copper-beryllium.

Light alloys are widely used for waveguides, waveguide components and television camera bodies, especially the magnesium alloys N4 and N6 which have good corrosion resistance and good mechanical

Development stage of thin-film production

properties. They are very suitable for machining by methods such as routing, and N4 alloy can be welded without much difficulty.

Satisfactory brazing is really only possible with unalloyed aluminium but this has poor mechanical properties. As there is no perfectly satisfactory solder for use directly on aluminium or its alloys, the problem of the low temperature joining of aluminium alloys is generally overcome by copper-plating the alloy, tinning the areas to be soldered, removing the exposed copper chemically and finally sweating the tinned surfaces together.

When weight conservation is important, as in television camera bodies, magnesium-rich alloys are very attractive, having a density of 1.87 compared with 2.87 for the engineering alloys of aluminium. Unfortunately, these alloys readily corrode if left in an unpainted condition even in mild environments, so they are not suitable for the manufacture of items containing moving parts that cannot be painted.

STAINLESS STEELS

There are two main types of stainless steel, austenitic and martensitic. Austenitic stainless steel is relatively soft, non-magnetic and very resistant to corrosion. This type is often used for domestic items.

Martensitic stainless steel can be hardened, is magnetic and though much more resistant to corrosion than ordinary hard steel is not as resistant to staining as the austenitic type. Table knives are made of martensitic stainless steel.

Both of these types are widely used in the electronics industry but in small quantities. The austenitic steels are particularly useful because of their non-magnetic properties, especially in the fully annealed, or mildly cold-worked condition, when they have the mechanical properties of mild steel.

Some of the special austenitic stainless steels have found use in unusually arduous conditions, e.g. type EN58J is used for equipment that is to be immersed at such a depth in the sea that oxygenation is only very slight. (Oxygenation is necessary for the prevention of corrosion in most stainless steels.)

The assembly of an r.f. module for Skynet II on the left. A completed module, with the cross-welds necessary for weight-saving, on the right





WAVEGUIDE COMPONENTS

Waveguide components are often difficult or even impossible to make from the normal rectangular tube. In such cases, recourse is made to electroforming techniques, using stainless steel formers.

When the shape is such that a steel former could not be withdrawn at the end of the electroplating operation, a thermoplastic former of the desired shape is made. After electroforming, it is removed either by melting or by dissolving out with a suitable solvent. Electroforming methods are slow and a single component may take several days to build up to the required thickness for adequate rigidity.

A recent innovation is the production of simple waveguide components, for example, tapers, by metal spraying techniques, zinc or silver-tin being the metals most often used. A thin deposit about 0010in thick is applied to the mandrel that has previously been treated with a suitable release agent. Before removing the mandrel, the coating is strengthened with resin-glass cloth laminates wet-laid on to the sprayed metal.

In an increasing number of cases it is essential that the dimensions of waveguide components remain constant in spite of varying temperature. This requirement is achieved by the use of alloys having a low coefficient of expansion, such as invar and nilo. The poor electrical conductivity of these alloys is overcome by electroplating with silver of sufficient thickness that all the radio frequency current is confined to the electro-deposit.

SOLDERABILITY

Alloys similar to invat are used for the leads of transistors and miniature valves in order to obtain the satisfactory glass-to-metal seals required by these components. This application gives rise to soldering problems, because these alloys are not solderable using non-corrosive resin-based fluxes.

In most cases the leads are gold-plated in the course of the transistor manufacturing process but, because the gold is readily soluble in soft solder, the joint is in fact made to the low expansion alloy.

Removing a high-power valve, weighing about 90lb, from a broadcast transmitter

The solderability of component leads was once a serious problem. The leads used to be made of copper wire electroplated with tin but, if the time between electroplating and soldering was more than a month, soldering became increasingly difficult, eventually reaching the stage when it was impossible. This trouble has been overcome in recent years by electroplating the copper wire with 70:30 tin/lead alloy which remains solderable for years, even under adverse conditions of storage.

The use of this alloy also overcomes another difficulty experienced when pure tin is used, namely that of the growth of tin whiskers. These are single crystals in the form of fine whiskers that grow out from the tin surface sometimes reaching lengths of lin. Such whiskers can cause short circuits in equipment. The growth of whiskers is inhibited by the presence of metals that alloy readily with tin, of which lead is the most convenient. A lead content of three per cent or more gives complete immunity.

PRECIOUS METALS

Among the precious metals, silver is used because of its high conductivity, both as an electro-deposit and on contacts as the block metal or in composites such as graphite-silver, sintered from powder.

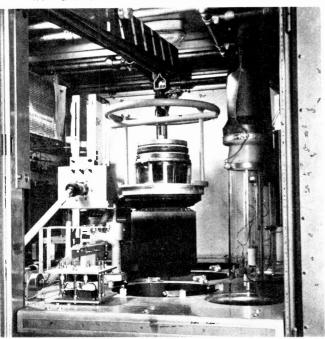
Gold is used as a non-tarnishing finish on copper and silver and in the form of modern "hard" golds, which may contain up to two per cent nickel or cobalt, as wear-resisting finishes for light to medium pressure applications.

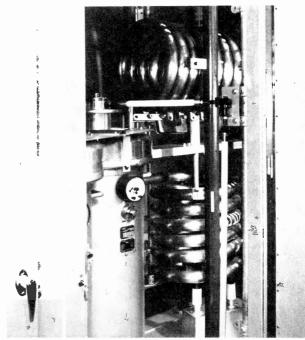
Rhodium is used for surfaces subjected to severe wear or as the opposing surface to gold in such bearing surfaces as slip-ring brush combinations, the slip-ring being gold plated whilst the brush is rhodium plated.

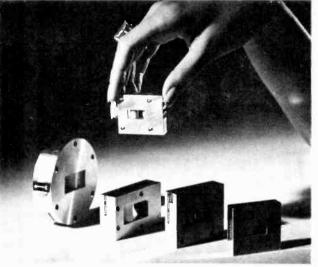
HIGH POWER VALVES

Whilst transistor devices have gone a long way in replacing small valves, the manufacture of large transmitting valves is expanding. The electrodes in these valves are made of high conductivity oxygen-

Gas-filled anode tuning capacitors with anode and output coils of 750kW broadcast transmitter







A series of isolators for use in waveguides

free copper and nickel whilst their supports are of refractory metals such as molybdenum, tantalum or tungsten sheathed in platinum.

All the metals must be of a high purity, in particular being free from gaseous impurities. Valve manufacturers experience great difficulty in obtaining refractory metals of sufficiently high purity.

NEW APPLICATIONS

Investigation into the behaviour of metals and other substances at temperatures well removed from ambience has disclosed properties not normally encountered, some of which are useful for specialised applications. Typical examples are the electrical super-conductivity of certain metals at temperatures near to absolute zero (-273 °C). It is possible that continuing research on these lines may have repercussions on the types of metal used throughout the electronics industry.

ACKNOWLEDGEMENT

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Marconi Mark VII colour television camera at the Leeds studios of Yorkshire Television



NEWS BRIEFS

MICRO-COMPUTER DEVELOPMENT

THE US space agency NASA, has selected RCA to develop a test model of a space computer that would be 100 times smaller and lighter than equivalent commercial systems.

The microprocessor could be the forerunner of a system for future manned and unmanned space vehicles such as the Space Shuttle and the Earth Orbiting Space Station.

The shuttle is a reusable craft that will take off like a rocket and land like an aeroplane. The space station would be a permanent orbiting laboratory able to accommodate a contingent of scientists.

Heart of the computer under development by RCA will be 15 large-scale integrated (LSI) arrays—one-eighthinch square chips each containing up to 600 electronic elements. Although the computer will weigh just 10 pounds, occupy one-half cubic foot and require only 15 watts of power, it will be capable of processing functions equivalent to room-size commercial computers.

CHANNEL SURVEILLANCE SYSTEM

THE first radar station to be ordered for the purpose of monitoring shipping movements in the Channel has been installed for the Department of Trade and Industry at H.M. Coastguard station at St Margaret's Bay, near Dover. A Decca 48 mile HR 729 system with two 16in displays, a 9ft scanner and 25kW transceiver, it will range from Boulogne to Dunkirk and over a similar arc of the Eastern approaches to the Dover Straits.

This is the direct outcome of the Government's concern for the accident rate in the Channel, which led to the National Physical Laboratory with the assistance of Decca, carrying out two major traffic surveys based at Dungeness and St Margaret's Bay, for the Department of Trade and Industry.

The new radar is the first ever sold to H.M. Coastguard and will also be used for search and rescue of vessels in distress when occasions arise. Its choice follows successful trials at Deal of a smaller Decca radar, which first caught the headlines when, in conditions of thick fog, the Deal Coastguard homed the Ramsgate lifeboat onto a yacht in distress and subsequently conned both back to Ramsgate Harbour.

EVENING LECTURES

A SHORT course of six evening lectures will be held at North Staffordshire Polytechnic, Stoke-on-Trent. Time: 6.30 to 9.0 p.m. All enrolment enquiries should be made to the Polytechnic.

January 12	M. Hughes	General Introduction to D.I.Y. Electronics
January 19	A. Douglas	Electronic Music Synthesis
		(including ref: to the P.E. Organ)
January 26	F. Hyde	Radio Astronomy
		(building a simple radio telescope)
February 2	A. J. Dunn	Radio Control of Models
February 9	A. J. Dunn	Electronics for Automobiles
February 16	M. Hughes	Logic Demonstration

All the above are regular contributors to PRACTICAL ELECTRONICS



WATER ON THE MOON

Till now, the examination of the rocks and dust brought back from the Moon by America and Russia, has indicated that the moon is entirely without water. Doubt is now thrown on these conclusions by Dr J. W. Freeman (chief *A pollo* programme investigator for the instruments) of Rice University, and a statement has been made about his reasons for deciding that there is water in pockets on the Moon.

The effect that this will have on the future of moon stations is very considerable for it would mean that the crew would not have to bring water with them from earth, or manufacture it from chemicals or re-cycling processes.

Dr Freeman and his associate Dr Kent Hills base their statement on the fact that the instruments left on the moon by *Apollos 12* and *14* detected a water cloud that was like a geyser in eruption. The vapour was very diffused and probably amounted to about a quart of water. The eruption was from a crack in the moon's surface and the event was recorded on instruments which are called suprathermal ion detectors.

The largest event that was detected covered an area of 10 square miles, and was in the region of the Sea of Storms. The accuracy of the exact position is not as good as could be hoped since there are only two instruments operating. However, the precise location can be established later.

One other point that supports the contention by Dr Freeman is that the geyser coincided with small moon quakes recorded by the seismographs left on the moon. The advent of cracks is therefore well supported.

Planetary scientists have always held that in the solar system planets all contain carbon dioxide and water in their interiors; with the geyser like eruptions on the moon is the proof of the fact. It will of course require further supporting evidence before acceptance but no doubt this will be dealt with on the remaining two missions.

ROCK HARVEST

The geophysical harvest of the *Apollo* 15 mission has been excitedly acclaimed by scientists. It will take a great deal of time to examine fully and analyse the rock and dust samples amounting to about 170lbs. They have a much wider variety of types and some 60 large pieces. The rocks which have been called Genesis are not as hoped the primeval lunar material.

One of the reasons for the choice of the Fra Mauro site on the *Apollo 15* mission was the hope that primitive finds would be made. However, the amount of information that will be obtained from the samples is very important.

HOT MOON

The moon is a hot body with the heat concentrated in the interior or alternatively in pockets of radioactive material. These pockets are perhaps around a hundred miles down from the surface.

This evidence of the heat gradient was obtained from the two probes which were put down to a depth of 6ft into the moon by the *Apollo 15* astronauts. The heat flow measurements showed that the increase in temperature amounted to about one degree farenheit for each increase of one foot in depth.

Compared with the earth this is about one fifth of the escape flow. This would indicate that there are radio-active minerals inside the moon. If the abundance is evenly distributed in the body of the moon then there must be a molten core. Melting temperatures would appear at about 300 miles below the surface. If this is not the case then the radio-active materials must be in pockets at about 100 miles depth.

MOONQUAKES

There are now on the moon three seismic stations which were set up by the crews of *Apolios 12*, *14* and *15*. The network has brought an immense amount of new knowledge of day to day happenings on and in the moon which were not even suspected. The happenings were disguised from previous observations because of the rigidity of the shell.

The power of some of the tremors would undoubtedly have made cracks in the earth's surface which is much less rigid than the moon. The major effects take place when the earth and the moon are near to each other each month. The forces that are set up due to the increased tidal pull and the effect of the changing barycentre cause the tremors which are transmitted back to earth. One such record showed that the disturbance originated some 500 miles below the surface which may be even greater than those occurring on earth.

The moon then is a body that is convulsed internally without visible signs on the surface.

CORE SAMPLES

A $8\frac{1}{2}$ ft core sample was taken from the moon and this was found to contain some 57 layers of soil. This illustrates about 2.400 million years of the history of the moon. The layers are of varying thickness and range from half an inch to five inches. It is possible that each layer represents a different meteorite impact.

At the bottom of the sample core a 3½ in section was of much coarser material containing some chips up to ½ in diameter. This could be basic rock which was overlaid by the rest of the material.

MYSTERIOUS MINIQUAKES

One of the mysteries of the seismic recordings of the moon are the swarms of minor disturbances quite unrelated to tidal effects.

The majority of the activity on the moon lies in a small area below the Ocean of Storms and this is where the deep tremors are recorded. These may be due to heat from the depths of the moon. The mini swarms however, cannot come from this source.

One suggestion is that they are triggered off by the mascons. These are areas on the moon where there is a specially dense region. These cause an out of balance effect and it may be that this is the answer. Nothing definite can be said about this at the moment. If this should not prove to be the answer then there must be some other activity that has not yet been detected.

The swarms of mini-quakes over a period of two and a half days showed four swarms and thirty events.

SATELLITE TROUBLE

There has been some trouble with the ATS-3 satellite, whose full title is Applications Technology Satellite-3. It is in a synchronous orbit at 70 degrees West longitude 22,300 miles above Colombia.

Fitted on this vehicle is an aerial which should rotate at 100 rev/min. This is necessary to keep it pointed at the earth while the satellite spins. About the middle of July 1971 the aerial began to slow up and fell to 80 rev/min and sometimes stopped altogether and sometimes revolved at other speeds.

It would appear that this is due to the sun which when it is north of the equator heats up the satellite and causes the aerial to seize up. When the sun moves south of the equator all should be well again.

Practical Electronics January 1972

Though not to be taken too seriously, this highly sensitive circuit provides a fascinating tool for psychological experimentation. Without any modifications it can be used as an audible pulse monitor.

The psychogalvanic reflex has long been used by experimental psychologists, policemen, and others, as a "window on the mind". Indeed a whole cult of fringe medicine has sprung up around it: self-styled "scientologists" spend hours questioning each other while holding the electrodes of a resistance meter. Now, thanks to the availability of integrated circuits, the enterprising constructor can make an instrument of much greater sensitivity than was possible until a year or so ago.

WHAT IS MEASURED?

If someone tells a lie they usually blush. This reflex, whose biological origin is not entirely explained, is associated with a number of small but detectable metabolic changes—respiration, heart rate and perspiration being the most easily measured.

Changes in perspiration can be detected almost instantaneously by the simple expedient of measuring the resistance between two electrodes placed on the skin. While one can observe some changes with something no more elaborate than an Avometer, one would be limited to subjects whose skin resistance was within the range of the meter, and one would certainly not be able to observe small changes such as occur with a passing thought in the mind of the subject.

In general a lie detector must register resistance changes as small as one per cent over a period of up to five seconds, the average value of resistance being anything between 10 kilohms and 10 megohms. As one is limited to a few volts at the electrodes (more would cause an unpleasant tingle), it turns out that a versatile instrument must detect current changes of a few nanoamps (amps $\times 10^{-9}$) per second. The instrument should also respond with equal sensitivity when the percentage change is the same, but when the actual current is a thousand times as great.

By O. COCKERELL

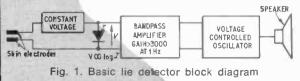
The block diagram, Fig. 1, shows how the huge range is encompassed in a single instrument without need for adjustment. The skin current is passed through a silicon diode which produces a voltage proportional to the logarithm of the current through it. Thus equal percentage changes of current will produce equal increments of voltage, about 2 millivolts for a 10 per cent change. The amplifier is in fact a differentiator below a few hertz so that an output appears only for a change of input.

HUM AND NOISE

Anyone who has put his finger on the input of a high gain amplifier will tell you that all you get out will be an almighty hum, and possibly some Radio 1 as well! For this reason components are included to provide 12dB per octave roll-off above a few hertz.

THE READOUT

Various readouts are possible, perhaps the simplest being a voltmeter. The present instrument, however, uses a voltage controlled oscillator and a crystal microphone as a speaker, so that a lie registers as a note of rising pitch. This serves two purposes: most people can hear a small percentage change of pitch over a very wide range, so that greater sensitivity than with a meter is possible; and secondly a sort of positive feedback occurs—as the subject hears the accusing little squeak he responds by making it even worse. It is also possible to use an earpiece if the subject is to be kept unaware of his responses.



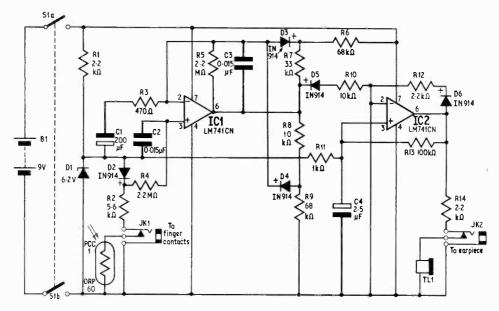


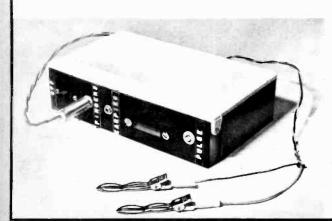
Fig. 2. Circuit diagram of lie detector

PULSE MONITOR

A bonus offered by the circuit comes in the form of a pulse monitor. If a photoconductive cell is used in place of the finger electrodes, and the subject's fingertip placed over it, his heartbeat can be heard in the form of frequency modulation of the output tone. The way it works is that the light passing through the finger is modulated in intensity by the blood pulsing through it.

THE 741 OPERATIONAL AMPLIFIER

If you have never used an operational amplifier before, or if you have fiddled for hours trying to get an earlier type to behave, only to find that it has succumbed to your ministrations and given up the ghost—take heart! The 741 is virtually indestructible. You can short the outputs to ground or to either supply and it doesn't mind. Even if you connect the battery the wrong way round the battery



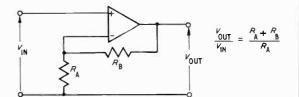


Fig. 3. Non-inverting operational amplifier configuration

will be the first casualty (but don't try this with anything bigger than a PP3).

If you are not quite sure what an "op.amp." does it is simply this: the output goes positive if the "+" input is positive of the "-" input, and goes negative if the opposite is true. The input currents are tiny, usually less than a tenth of a microamp, and the output resistance is low, a few hundred ohms. The open loop gain—that is without feedback—is tens of thousands as long as the inputs are not allowed to get a sniff of the output in high impedance circuits.

CIRCUIT DETAILS

Fig. 2 shows the complete circuit diagram. An integrated circuit, IC1, is used as an amplifier in the standard non-inverting configuration. Fig. 3 shows this in simplified form. The gain is simply

$$\frac{R_{\rm A}+R_{\rm B}}{R_{\rm A}}$$
 or if $R_{\rm B} \gg R_{\rm A}$, simply $\frac{R_{\rm B}}{R_{\rm A}}$.

In the actual circuit $R_{\rm B}$ is 2.2 megohms and $R_{\rm A}$ is 470 ohms so the gain is about 4,000 at around 1Hz. The capacitor C1 starts reducing the gain below 1Hz, and C2 and C3 reduce it above 3-4Hz.

COMPONENTS

Resistors

- 2.2k () R1 R2 5.6k Ω **R**3 **470** Ω
- R4 2·2Μ Ω R5 2·2M Ω
- R6 68k Ω
- **R**7 33k Ω
- $10k \Omega$ R8
- R9 68k Ω
- R10 10k Ω
- R11 $1 k \Omega$
- R12 2.2k Ω
- R13 100k Ω R14 2·2k Ω
- All $\pm 5\%$, $\frac{1}{4}$ W carbon

Light Dependent Resistor

PCC1 ORP60 or U1100 Miniature CdS photocell

Capacitors

- C1 200µF elect. 4V
- C2 0.015µF
- C3 0.015µF
- C4 2.5µF elect. 64V

Integrated Circuits

IC1, IC2 LM741CN, SN72741 or 741-OPA (8-lead D.I.P.) (2 off)

Diodes

D1 6.2V 400mW Zener D2 to D6 IN914 or any small silicon signal diode (5 off)

Speaker

TL1 Crystal mic. insert

Switch

S1 Double pole changeover slide switch

Jack plugs and sockets

- JK1 3.5mm Jack socket with break contact, and plua
- JK2 2.5mm Jack socket with break contact, and pula

Miscellaneous

Crystal earpiece (optional) Plastic box 6in × 4in B1 9V Battery PP3 Sprung hair clips Veroboard 0-1in matrix, 5-7in imes 2-8in with copper strips Battery connector Capacitor clip (for mounting TL1)

The diodes D3 and D4 together with R6 to R9 reduce the gain as the output swings above zero and below -2 volts. These voltages are with reference to the positive end of the Zener. Their purpose is to reduce the time taken to recover from a very large input swing.

VOLTAGE CONTROLLED OSCILLATOR

Fig. 4 shows the essential components of the voltage controlled oscillator. The 100 kilohm and 1 kilohm resistors apply positive feedback so that small voltages at the input cause the output to swing to its limits (about +2 and -4 volts, with respect to the positive end of the Zener). One hundredth of the output appears at the "+" input by potential divider action.

Assuming the output has just switched to its upper (+2V) limit, diode $D_{\rm F}$ will be forward biased and the "+" input will be at +20 millivolts. Capacitor $C_{\rm T}$ will charge up through $D_{\rm F}$ and $R_{\rm F}$ until its voltage reaches the 20 millivolts on the "+" input when the output swings rapidly to -4 volts, taking the "+" input to -40 millivolts. Diode D_F is now reverse biased and $C_{\rm T}$ can only discharge through $R_{\rm T}$ at a rate dependent on the input voltage $V_{\rm C}$, which is assumed to be at a negative potential. As soon as the "-" input reaches -40 millivolts, the output switches to +2 volts and the cycle starts again.

Thus the output is a series of pulses of almost constant width but of repetition frequency determined by $V_{\rm e}$

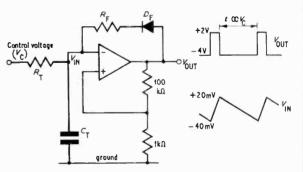


Fig. 4. Voltage controlled oscillator circuit

THE SKIN ELECTRODES

The finger clips or skin electrodes are the only items unlikely to be found in the average constructor's inventory-they are in fact sprung hair curlers. Get the nickel-plated type rather than the aluminium ones which are difficult to solder. I got mine from Woolworths.

MANHATTEN LAYOUT

Layout of components and cuts in the copper strips are shown in Fig. 5. The form of Veroboard layout used will probably seem wasteful at first

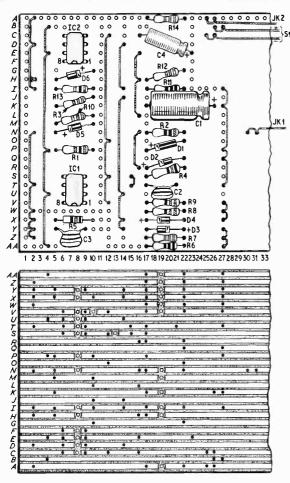


Fig. 5. Component assembly and wiring details of Veroboard panel (Note that the copper strip should be cut at 19T)

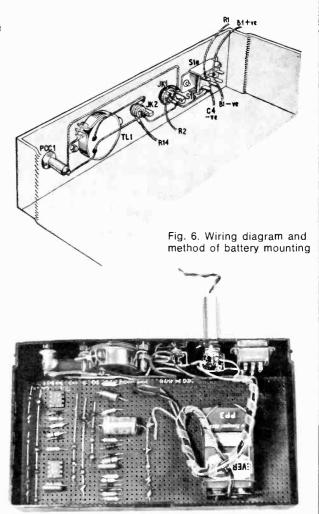
glance, but it has the advantage of being easy to implement and to check. All components are more or less parallel to the copper strips, while component side links run at right angles (like the streets of Manhatten). The battery slides into three 20 s.w.g. straps on the board which prevent it from moving about, but allow it to be replaced (see Fig. 6).

Apart from the battery clips only four wires connect to the board—input, output and positive and negative supplies. The negative supply is the common rail for input and output.

The miniature jacks used for input and output are of the type which break a connection when the plug is inserted.

WARMING UP TIME

When you first turn the machine on, the large electrolytic, C1, may take a minute or so to polarise —that is, as well as taking a few seconds to charge up, and it can take considerably longer for the leakage current to reduce to its proper value. The pitch of the oscillations should fall to a very low value,



certainly less than a hundred hertz, before asking questions. A simple test to establish that the device is "in range" is to get the subject to take several deep breaths, this should elicit a full scale response.

OPERATION

Put the finger clips on the subject with soothing reassurances that he won't get electric shocks. Then proceed to shock him by proving that the gadget works. A simple demonstration is to ask him to think of a number, say between one and five, and then ask him about each number in turn. He must of course say no to each number so that he must lie once. Present the numbers in random order otherwise the reaction as his number approaches might be confused with the previous one.

WARNING

Don't try to operate the gadget from a mains supply, or battery eliminator, even small leakage currents can be felt, and larger ones could be dangerous.



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MAKING TRANSISTOR RADIOS, A BEGINNER'S GUIDE, by R. H. Warring, £1.10. Postage 10p.





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STATE OF THE INDUSTRY

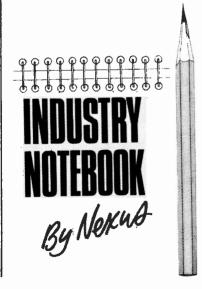
Official statistics necessarily trail events. It takes time for companies to make their returns to the Department of Trade and Industry and more time to analyse and collate them. So we had to wait until November 3 before getting the 1971 first half results on how the capital equipment sector is making out in these troubled times.

The figures were not as bad as some had feared. Compared with the first half of 1970 total output was up 5 per cent at £336.6 million. Radar was a bright feature. up 28 per cent, but slumps in broadcast equipment (--20 per cent), navigational aids (-11 per cent), and radio communications equipment (-10 per cent) eliminated the gains. Control equip-ment put on 22 per cent and nucleonic instruments a cracking 25 per cent, although the latter is such a small segment of the industry that the gain had little effect on the overall totals. Exports were down from 40 per cent to 36 per cent. Surprisingly, employment remained constant, this year being 150,000 people at the end of both the first and second quarters.

The red light is seen in the forward order book. The figure at June 30, 1970, was £648 million. On the same date in 1971 the figure had slumped to £572 million. The conclusion must be that the great days of 15 per cent growth every year are all but over.

LEANER AND TOUGHER

Squaring up to the commercial facts of life is no easy matter. Some people throw in the sponge as did RCA with computers. Since then the U.S. General Electric has pulled out of microcircuits. After



an investment of £20 million and four years of struggle.

Others persist by seeking economies and building better management structures. Plessey recently underwent a complete reorganisation converting a cumbersome operation into "businesses", each accountable directly to the main Plessey board.

Some details of how Mullard tackled their problem have been revealed by commercial director Jack Akerman. Mullard now has four divisions each responsible for a market sector. They are Consumer, Computer, Communications, and Instrumentation and Control. The general manager of each division looks after his market sector not only from the sales point of view but also the Mullard manufacturing plant producing goods for his market area. Each is charged with making a profitable return on the capital employed which totals some £90 million right across the board.

Everyone is expected to work for the benefit of the whole company—not just for his own division. Akerman regards the organisational change as a platform for growth in the '70s. Results after 6 months showed a reduction in stocks of over 10 per cent while providing a better service to the customer, reduction in the commercial staff and better morale.

SLIMMING DOWN

Another example of slimming down to better meet the future as well as today's problems is to be seen in SGS whose U.K. operations are based on a marketing unit at Aylesbury and a substantial manufacturing plant at Falkirk. Fears that the Falkirk plant would be closed entirely were recently allayed by a lightning visit from Milan by Giancarlo Maimone, newly appointed managing director, who had discussions with local MP Harry Ewing and union representatives. The outcome was that Falkirk appears to be saved, for the time being at least, but with a labour force pruned from a peak of 1,100 down to some 500-600.

The marketing team at Aylesbury has also been streamlined and it is possible that a joint marketing team with ATES (a company associated with SGS through the Italian holding company IRI) will be set up. Other SGS plants in Italy, France, Germany and Singapore are all reported to be working well under capacity. Mr Maimone has some tough decisions ahead in relation to cut-backs elsewhere but, for the moment, SGS men in the U.K. are optimistic. "At least we now have a clean situation on which to build", was the opinion of marketing manager Roy Hood.

TELECOMMUNICATIONS BOOM

One sector of the industry which is flourishing is telecommunications which, with the introduction of PCM and electronic exchanges, is getting more and more electronic in character. The TXE 2 electronic exchange is now firmly established. Since the first was commissioned in Ambergate, Derbyshire, in 1966, some 150 have been installed and they are now coming into service at a rate of more than two every week.

The big TXE 4 with up to 40,000 lines (TXE 2 has only 2,000 lines) has now been finally ordered by the Post Office from Standard Telephones and Cables. The development of TXE 4 has not been without problems but these have now been resolved. First contract is worth £15 million and the first installation will be made in London at the beginning of 1975. Soon afterwards, TXE 4's will be in service in Manchester and Birmingham.

The TXE 2 has been a big success for Plessey who have supplied the bulk of them to the Post Office and have also been successful in selling them overseas.

But the biggest telecommunications boom area of all is data communications. Fifteen countries in Europe have now come together in sponsoring a year's study by PA International Management Consultants in London on the data communications requirements of Europe for the next 15 years. The study is planned to show future customer requirements, anticipated data traffic flow with Europe, and between Europe and other continents, and future developments in data processing systems.

BRITAIN IN SPACE

The biggest rocket to get airborne (but only just) on Guy Fawkes day was *Europa* 11 launched from Kouro in Equatorial French Guiana. Regrettably, it was the biggest flop of the year.

A few days earlier the British technology satellite *Prospero* (formerly X3) was successfully launched in orbit by a *Black Arrow* from Woomera and has been functioning well.

Prospero's success, however, will not compensate us in prestige for the loss of Europa 11. We pulled out of the European Launcher Development Organisation (ELDO) in 1968. It was a controversial move which did not endear us to our continental partners. The Blue Streak rocket used on November 5 was sold to ELDO. One crumb of comfort was that Marconi's navigational guidance system was not at fault.



THE ARLINGTON DICTIONARY OF ELECTRONICS

Published by Arlington Books 171 pages, $8\frac{1}{2}$ in \times $7\frac{1}{2}$ in. Price £3

This new book of American origin is well produced and attractively presented. The range of material covered is enormous, ranging from simple circuit theory to microwave engineering and quantum physics. Also included is some advanced maths including vector calculus. Digital computer fundamentals are also dealt with. Diagrams, though not profuse, are clear and well explained and the definitions are succinct and, wherever possible, assume only elementary knowledge.

My main criticism is not of this book itself but of the whole philosophy of presenting technical information in dictionary format. It is difficult to imagine when this type of book would be of use. In text books obscure terms are usually given lengthy explanations whilst a lay reader encountering technical terms would be unlikely to learn much from this type of book.

However, if this type of book appeals to you I can recommend it as an addition to library or personal book collection.

S.R.L.

110 INTEGRATED CIRCUIT PROJECTS FOR THE HOME CONSTRUCTOR

By R. M. Marston Published by Iliffe 128 pages, 8½in × 5⅔in. Price £1.20

N cramming 110 circuits into as many pages Mr. Marston has undoubtedly achieved quite a feat. The integrated circuits described range from the CA3018, which is simply four transistors on one chip. to the PA246 five watt amplifier. There is also a section describing 35 RTL digital i.e. projects.

The general procedure used is that a description of the internal circuit of the i.c. is presented, then several projects are described so that the reader becomes familiar with the function of the unit. Many of the projects are useful and interesting and it is almost certain that everyone will find something of use in this book.

The presentation of the information is, however, not all that good. Each new project does not have a heading so one has to resort to the index to find a particular circuit. All the i.c.s, apart from the digital, are represented with the operational amplifier symbol of a triangle. For i.c.s such as the CA3018 this makes it very difficult to read circuit diagrams, and on some diagrams output signals seem to come out of inputs.

At one new penny per circuit the book is certainly good value but a little more care in presentation would have helped.

R.S.

QUESTIONS AND ANSWERS ON RADIO AND TELEVISION (3rd edition) By H. W. Hellyer

Published by Newnes-Butterworth 172 pages, $6\frac{2}{4}$ in $\times 4\frac{2}{4}$ in. Price 60p

This book is clearly aimed at people who would like to know how television and radio systems operate, but whose knowledge of electronics is slight. It takes the reader through the main principles of electronics and then goes on to explain, in simple terms, the basic features of electronic devices, including the valve and the transistor. Some of the important circuits in radio and television are then described with many circuit diagrams to illustrate practical considerations. The final chapters deal with the functions of the television receiver, including tuning, synchronising and picture linearity control. Colour reception is not dealt with in any detail.

The sections on basic electronics are lengthy and, on the whole, they provide a useful introduction to the subject for a beginner. However, in a book with this title, I think that some of the questions and answers on semiconductor manufacture would have been better replaced by explanation of some practical points, such as the methods of correct biasing for transistors. The description of the field effect transistor, which is anyway complicated by the use of the term "depletion layer" without definition, seems superfluous to this kind of book.

Apart from these criticisms the book provides a lot of useful information on radio and television and would give the interested beginner plenty of food for thought. S.R.L.

RADIO AND TELEVISION YEAR BOOK 1971/72 Edited by Eric Ickinger

Published by IPC Electrical-Electronic Year Books Ltd

224 pages, 8 $\frac{1}{2}$ in \times 5 $\frac{1}{2}$ in. Price £1

A COMPLETE guide to radio and television broadcast stations, transmitting at v.h.f. and u.h.f. in the U.K., precedes the up-dated cataloguing of commercial equipment under classifications: monochrome television, colour television, radiograms, record players, table radios, portable radios, car radios, tape recorders, unit audio.

The format follows the usual style as a buyer's guide giving descriptions, brief specifications, and prices (November 1971). Some illustrations are also included.

HI FI YEAR BOOK 1972

Edited by Colin Sproxton

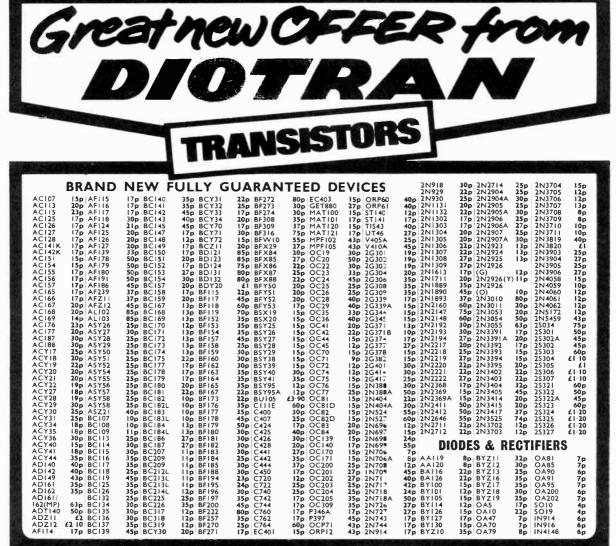
Published by IPC Electrical-Electronic Year Books Ltd

376 pages, $8\frac{3}{4}$ in \times 6 in. Price £1.25

A MONG the preceding articles in this latest Hi Fi Year Book is an interesting outline of the Dolby noise reduction system that is now being increasingly incorporated in commercial equipment. Also included are: Hi Fi from FM radio; a short review of eighteen of the best recordings over the past year or so: the meaning and importance of pickup parameters; a glossary of terms associated with microphones; a list of common abbreviations and meanings.

The equipment descriptions and prices are updated but do not include all new items that took their bow at the 1971 Audio Fair (see our Audio Fair report in this issue). A complete list is given in the centre of this book of September 1971 list prices and new purchase tax rates.





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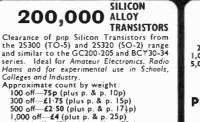
Price: 500 £9; 1,000 £15 TYPE ST N5. Silicon Planar Transistors npn TO-5 Metal Can. Types similar to: BFY50-51-52 and Metal Can. 2N2192-92.

Price: 500 £9.50; 1,000 £16

TYPE STPL. As above but in pnp and similar to types 2N5354-56, 2N4058-2N4061 and 2N3702-3. Also used as complementary to the above npn devices type STNL. Price: EOD (TIS): 1.000 (13) Price: 500 £7-50; 1,000 £13

TYPE STNK. Silicon Planar Plastic Transistor npn with TO-18 pin circular lead configuration, I.C. 200mA, 300mW and similar to BC1078-89, BC170, BC173, BC182-184, BC237-89 and BC337-8. Price: 500 69:50; 1,000 £16

When ordering, please state type required, i.e., STNK or STN18, etc.

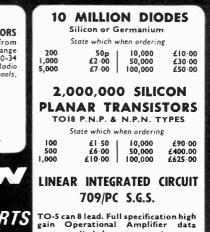


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www.i.c.audio mixer

Combines Mic or Tuner inputs into a single channel

By M. J. Bunce

This article describes a simple three channel, audio mixer using an integrated circuit preamplifier. It has a substantially flat frequency response in the audio range, and is designed for use with both domestic and semi-professional tape recorders, particularly with regard to cine soundtrack recording.

The mixer can also be used by pop groups, using impedance matching transformers if low output impedance transducers are used, such as dynamic or ribbon microphones.

CIRCUIT DESCRIPTION

The circuit of the mixer itself is given in Fig. 1. The input lines to the pre-amplifier are blocked to d.c. by capacitors C1, C2 and C3. This is necessary to prevent crackling and hiss caused by current flowing through the potentiometer wiper circuits.

The pre-amplifier is an integrated circuit operational amplifier contained in a standard eight-lead TO5 package. Such amplifiers are beginning to appear quite cheaply on the surplus market. The circuit described was designed around an S.G.S. μ A709, which has an open loop gain in the order of 50,000 at frequencies up to around 1MHz. The gain of the amplifier for a single channel is given by:

$$G = \frac{R_F}{R_I}$$

where $R_{\rm F}$ is the value of feedback resistance R5 in ohms and $R_{\rm T}$ is the value of input resistance. In this case:

Gain G =
$$\frac{1,000,000}{100,000} = 10$$

Thus the amplifier has an overall gain of ten.

CAPACITOR ROLES

To prevent the amplifier from amplifying stray radiation, picked up by the external connecting leads, the input is shunted by a resistor/capacitor network, consisting of R6 and C4. Similarly, generation of radio frequency parasitic oscillations within

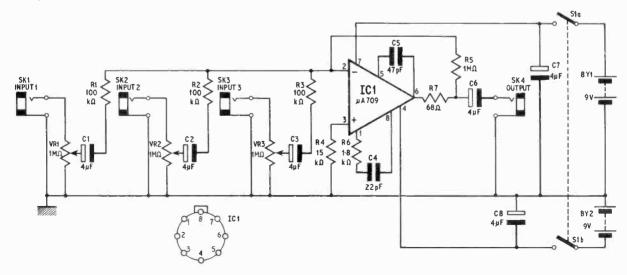


Fig. 1. Circuit diagram of audio mixer

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I.C. AUDIO MIXER

COMPONENTS .

Resistors

R1	100k Ω
R2	100k Ω
R3	100k Ω
R4	15k Ω
R5	1M Ω
R6	1-8k Ω
R7	68 Ω
All	$\pm 10\%, \frac{1}{4}W$ carbon

Potentiometers

 $\begin{array}{lll} \mbox{VR1} & \mbox{1M}\ \Omega \ \mbox{carbon} \ \mbox{log}. \\ \mbox{VR2} & \mbox{1M}\ \Omega \ \mbox{carbon} \ \mbox{log}. \end{array}$ VR3 1M Ω carbon log.

Capacitors

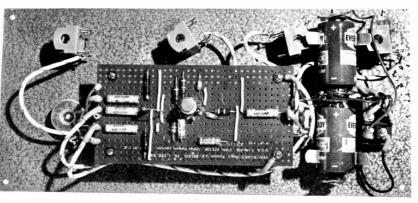
- C1 4μ F elect. 15V C2 4μ F elect. 15V
- C3 4µF elect. 15V
- C4 22pF polystyrene
- C5 47pF polystyrene
- $\begin{array}{rcl} \mathsf{C6} & 4\mu\mathsf{F} \text{ elect. 15V} \\ \mathsf{C7} & 4\mu\mathsf{F} \text{ elect. 25V} \end{array}$
- C8 4μ F elect. 25V

Integrated Circuit

IC1 µA 709 (S.G.S.) or equivalent

Sockets

SK1-SK4 Standard type jack sockets (4 off)



Switch

S1 Double pole on/off switch

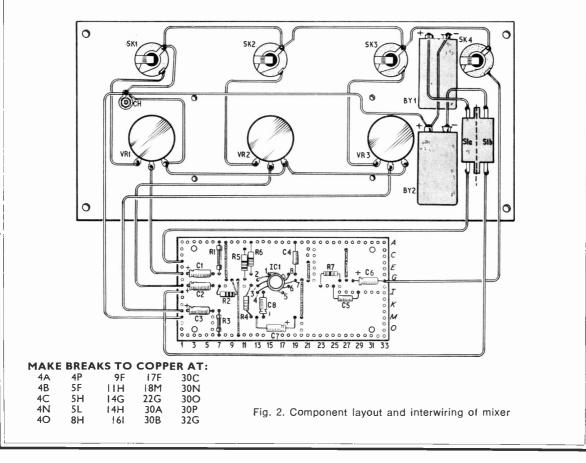
Batteries

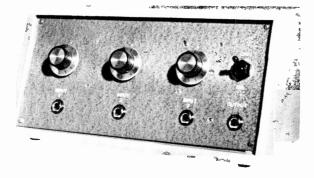
BY1, BY2 9 volts type PP9 (2 off)

Miscellaneous

Aluminium panel 16 s.w.g. 11in imes 5in Skirted control knobs numbered 0-10 over 300 degrees

Veroboard 5in \times 2½ in 0.15in matrix Hardboard and wood as required





the amplifier is suppressed by capacitor C5 at the amplifier output.

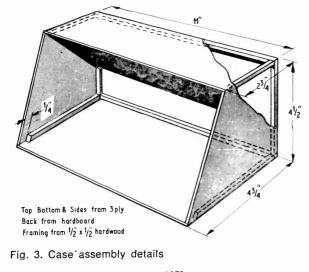
The values of these components determines the upper frequency response limit of the amplifier; its lower limit is determined by the values of the input capacitors C1, C2 and C3, and also the output capacitor C6. For this mixer application, a fairly sharp roll-off is required above 25kHz.

Any d.c. component in the amplifier output (e.g. that due to offset or drift within the i.c.) is blocked by capacitor C6, which also protects the amplifier against short-circuit fault conditions at the input stages of the equipment to which it is connected. It may seem at first that using the component values specified voltage offsets might easily occur. In practice, however, although offsets of up to $\pm 1V$ can occur at the output of the amplifier, this is of no consequence since the output is d.c. blocked, and the output voltage swing never exceeds $\pm 0.5V$.

INPUT INTERACTION

It would also appear that the inputs might interact (i.e. turning one input to minimum would cause a drop in gain of the remaining inputs). This is also not true, since the input of the i.c. is a virtual earth, and the inputs do not therefore have any effect whatsoever on each other.

As the power requirements of the amplifier are small, it was decided to use two 9V batteries, particularly as the amplifier requires a centre-tapped



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supply. The design of a suitable power unit for such a circuit could be disproportionately costly and complex. Besides, for the particular application of this mixer, portability is useful.

Earthing is not critical, although the fewer the number of separate earths, the better.

CONSTRUCTION

The unit is built around the front panel, thus avoiding numerous trailing wires, so that it may be fitted into the case as a complete module. Component assembly details on the front panel are shown in Fig. 2.

The electronic components are all mounted on a piece of Veroboard which is spaced from the panel by four 6BA spacers. The integrated circuit package is mounted on an i.c. socket to obviate the risk of damage when soldering the leads.

The layout of the components on the panel is not critical, however all unused whole copper strips should be electrically bonded to earth to avoid pickup of extraneous signals.

When laying out the front panel (see photograph), consideration was given to the ergonomic aspects of the design, particularly as the cine sound track operator could well be working outdoors in cold weather. As he could be wearing gloves under such conditions, plenty of room was left between the control knobs, which are of skirted type.

The case is of thin plywood, angled to provide a comfortable degree of rake to the front panel. Since each individual constructor may wish to design his own case to suit his requirements, only a simple perspective view of the prototype case is included (Fig. 3).

HUM SUPPRESSION

The inside of the case is lined with aluminium foil to prevent pick-up of external interference such as mains hum. The foil is earthed to the front panel at the chassis solder tag CH (Fig. 2).

The screws fitted to the front panel are all countersunk, with the exception of the four case securing screws, as this improves neatness and avoids any risk of fouling the skirt of at least one of the potentiometer knobs.

USING THE MIXER

In setting up the mixer each number on the control knobs represents a gain increase or decrease of approximately 1dB with the input and feedback resistance values given in Fig. 1.

For amplifier gain variations the input resistors R1, R2 and R3 can be changed either singly or totally to suit individual requirements. Thus, with the 1 Megohm feedback resistance given and input resistors in channels 1, 2 and 3 having values of 1 Megohm, 100 kilohms and 10 kilohms respectively, these will provide suitable gains for an f.m. tuner, crystal microphone and dynamic or ribbon microphone providing the latter two are fitted with impedance matching transformers.

No master gain control is fitted since for its particular design application the tape-recorder input gain control is satisfactory. Monitoring is carried out in the normal way using the recorder facility. \bigstar



Mowing by Moonlight

R ICH. insomniac Americans can now while away the midnight hours by mowing the lawn, without fear of disturbing their neighbours.

The Elec-Trac is a quiet and compact electric tractor developed by General Electric for use in the garden. It can mow, till, blow away snow, or provide a mobile power source for electrical equipment.

A programmed starting system prevents the operator from applying too much speed too soon and when the controls are set to "full forward" a built-in safety circuit automatically accelerates the vehicle smoothly to the desired speed.

Six long-life batteries will provide eight to ten years of life under normal conditions.

The company predicts a great future for after dark lawn care. They say, "We know that from three years of testing at night that quiet mowing in the cool of the evening or early morning is a real pleasure."



Electronic Cash Register

To MEET the need for faster check-out in supermarkets and stores, a new electronic system has been designed to replace the slow and cumbersome cash register. Pitney-Bowes call the new system "SPICE" (Sales Point Information Computing Equipment), and the unit which the customer sees is closer to a computer terminal than a cash register.

By connecting all the units to a central in-store controller, all information concerning sales can be recorded on magnetic tape. This can later be processed by a computer and any facts and figures which are needed can be extracted for use by the store manager. As well as accurately recording sales, the terminals may be used to get immediate notification of a customer's credit standing. The terminals can also be used as calculators to work out quantity prices or discounts.

An optional extra to SPICE is "PEPPER" (Photo-Electric Portable Probe/Reader) which eliminates the need for the sales clerk to ring up the prices. All sales items are individually marked with specal tickets, on which the price is coded into a row of bars. The electronic "pen" is quickly passed over the ticket, instantly displaying price, item number, and department number on the terminal register. Speed and elimination of human error make this very attractive to both customer and seller.

For small stores or places where access to a central computer is not feasible, a self-contained system known as "SALT" (Stand Alone Terminal) is available which can be programmed to provide on-the-spot point-of-sale information.





A Sight for Sore Ears

WHEN noise exceeds a certain level it is not only annoying, it can be positively dangerous! In an attempt to lessen the risk to hearing for people who live or work in noisy environments, the Noise Abatement Society is now marketing two new pieces of equipment: a survey meter which lights a lamp if sound exceeds the danger level and ear defenders for protection when it does.

Eye Movements Control Wheelchair

F OR patients who have lost the use of both arms and legs, this wheelchair offers new hope. The only actions needed to start, steer or stop the wheelchair are movements of the eyes. The sight-guide steering mechanism used in the chair was developed originally for NASA, the U.S. space agency, which is seeking ways of freeing the hands of pilots and astronauts for other in-flight duties.

A prototype wheelchair controlled by eye movements is at present being tested at the Institute of Rehabilitation Medicine in New York. If it is eventually put into production, an estimated 100,000 quadriplegics (those with no use of arms or legs) could be made mobile.

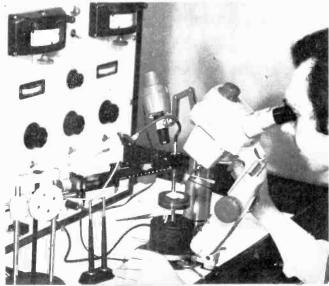


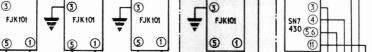


Microstrip Production

These new microwave techniques are being applied in many areas including the development of an experimental digital radio link for the Post Office. This will be designed to transmit and receive digital signals at 120 million bits per second.

The photo shows part of the assembly process using a vacuum probe micro-manipulator.





(4)

IC7

any value for C1. There is no critical lower value for the minimum signal pulse length. The integrated circuits IC1 to IC7 (Type FJK101 or 74121) are monostable multivibrators, each triggering the next in turn via the R-D-C network from pin I (\overline{Q} output), providing the interval period of about 0.32ms. The resistors in these networks (R2, R4, etc.) should not exceed 200 ohms. The sync pulse

The particular advantage of Coder 2 is the flexibility of timing since the timing resistance combination of R1-VR1 can vary between 1.5 kilohms and about 40 kilohms (R1 is fixed at 1.5 kilohms) with any value for C1. There is no critical lower value

ODER 2 is described in this article in two variants.

PART TWO

Although more expensive to make than Coder 1 described last month, they are easier to assemble. Both variants use the same basic printed circuit pattern apart from a minor change to the circuitry involving IC7.

CODER 2A

25

1 0

A

3

6

IC1

FJK 101

R2,1800

+1 CG83

Fig. 10 shows the circuit of Coder 2A. The grey panel area is the part which is altered to make Coder 2B. More about this later.

CODER 24

1.5 k0

(II)

(4)

IC2

FJK 101

Œ

4.180 Ω

CG83

/R 2

25 kΩ

Practical Electronics January 1972

+5V

OUTPUT

1 1

IC82

1

R14 1800

Use integrated circuits to simplify control functions

LOGICAL RADIO CONTROL By A.J. Dunn

VR5

25

60

1 10

Ð

R8,1800

04

CG83

IC5

25 k0

10

(14)

R10,1800

CG83

IC6

1.5 kΩ

25

>_{c7}

llŀ

0-2µF

IC4

(1) (10)

(14)

1.5 kΩ

0.2 0

IC3

FJK 101

14

3)

60

1

R6,180Ω

DI

CG83

COMPONENTS . . .

CODER 2A

and the second

Resis	sturs							
R1	$1.5 k \Omega$	R5	1·5kΩ	R9	1·5kΩ	R13	$22k\Omega$	
R2	180Ω	R6	180 Ω	R10	180Ω	R14	180 Ω	
R3	1·5kΩ	R7	1√5kΩ	R11	1√5kΩ			
R4	180Ω	R8	180Ω	R12	180 Ω			
All	$\pm 10\%$	wat	t carbon					

Capacitors

a latara

C1	0·22µF	C5	0·22µF	C9	0.22µF	C13	1.0µF
C2	2•2µF	C6	2 2µF	C10	2•2µF	C14	2.2µF
C3	0∙22µF	C7	0.22µF	C11	0.22µF		
	2 2µ F						
	tantalum						

Potentiometers

VR1 to VR6 25kΩ linear carbon (6 off)

Integrated Circuits

IC1 to IC7* FJK101 or SN74121N (7 off) IC8 FJH101 or SN7430N

D1 to D7 CG83 or OA10 (7 off)

Miscellaneous

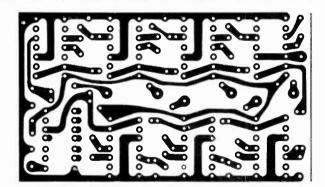
Fibreglass base printed circuit board and etching kit

is taken from the \overline{Q} output of IC7, the pulse length being 0.7CR or approximately 14ms.

For timing capacitors (C1, C3, etc.) of 0.2μ F and control resistance of 25 kilohms in series with 1.5 kilohms, the corresponding maximum signal pulse length is approximately 5ms.

One of the NAND gated inputs (pin 3) is connected to the common "earth" line except in the case of IC4 where an initiating switch is inserted to start the cycle of operation.

The output from the Coder is taken from IC8, a single eight-input positive logic NAND gate.



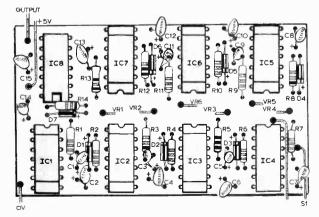
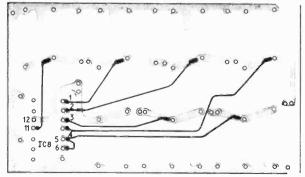


Fig. 10 (far left). Circuit diagram of Coder 2A. The area for modification to convert to Coder 2B is indicated by the grey area

Fig. 11 (left). The printed circuit pattern (full size) for Coder 2A

Fig. 12. Component layout of Coder 2A. The area where modification is required to convert to Coder 2B is shown in Fig. 16

Fig. 13 (below). Wire link connections on the copper side of the board



CODER 2A P.C. BOARD

The pattern for the printed circuit, which should be of fibreglass base, is shown full size in Fig. 11. As with Coder 1 described last month, it is worth providing an extra blank area of board at one end for mounting; this would add about $\frac{1}{2}$ in to the length of the working pattern.

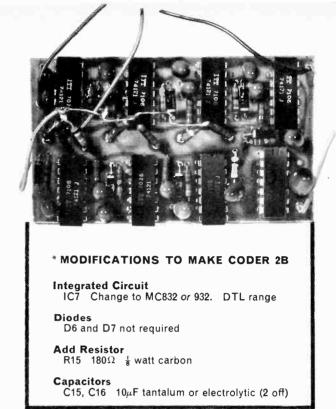
Component layout is shown in Fig. 12, while details of wire link connections are in Fig. 13.

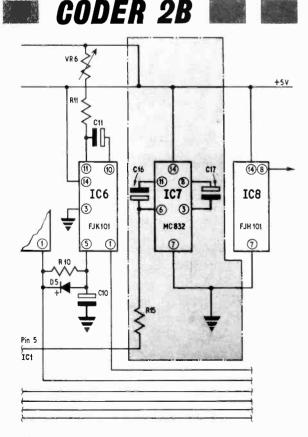
Orientation of the integrated circuits is important; if an error is made after soldering *in situ* it can prove a messy job to remove the i.c. unless a special desoldering tool is available.

CODER 2B

Coder 2B is similar to Coder 2A except it has a fixed cycle time. The integrated circuit IC7 is replaced by a dual four-input NAND gate (DTL type MC832 or 932) equipped with "node" inputs. The outputs from IC7 are capacitor coupled to the opposite "node" inputs, so forming a multivibrator, Fig. 14.

The output from pin 6 initiates the cycle for each period τ_1 (see last month). This period can be adjusted by altering the values of the coupling capacitors C16 and C17. For small variations a series resistor-potentiometer arrangement, as in the other stages, can be connected from pin 3 to the positive line and also from pin 11 to the positive line.





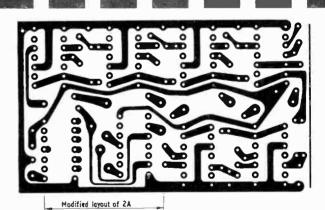
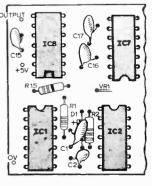


Fig. 14 (left). Modification of Coder 2A circuit from part of Fig. 10 to convert to Coder 2B

Fig. 15 (above). Printed circuit pattern (full size) for Coder 2B

Fig. 16 (right). Modified area of layout for Coder 2B



The modified printed circuit pattern is shown full size in Fig. 15, while the modified part of the component layout is shown in Fig. 16.

DECODER

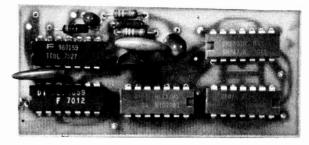
A decoder is essentially a means of translating the information transmitted in pulse code to some form of sequential mode analogous to the operation of model functions.

In the decoder described here, three dual JK flipflops make up a shift register (Fig. 17) driven by a negative going pulse, derived by feeding the input signal through two paralleled NAND gates. A retriggeragle monostable is also used to detect the signal period.

The circuit shown in Fig. 17b is used to produce the negative going pulses required by the shift register inputs to 'clock' and 'clear': this should be considered in connection with the waveform diagram Fig. 18.

The input signal (Fig. 18a) is applied to two parallel NAND gates, each with a "fan-out" of eight, producing the requisite negative going signal (Fig. 18b) adequate to drive the $6 (\times 2)$ clock input loads.

The 0.1μ F capacitor is used to slow the fast edges slightly and prevent spurious ringing effects. This signal is applied to the input of the retriggerable monostable whose output (Q) is maintained high (Fig. 18c) till after the last trigger (pulse No. 6)

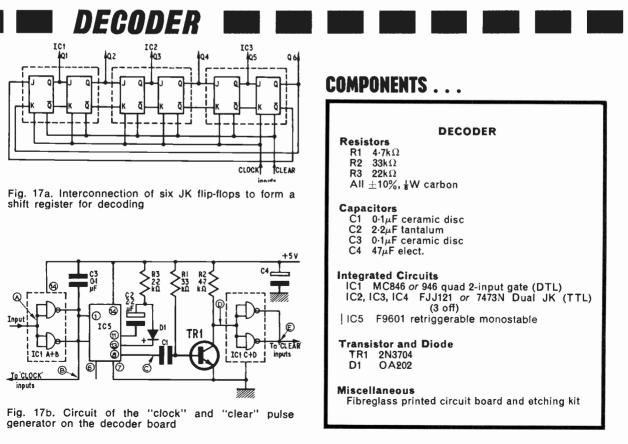


Finished decoder board

dependent upon C1 and R1 ($2\cdot 2\mu$ F and $20k\Omega$) as pulse width $\tau \simeq 0.36 R_1 C_1 \left(1 + \frac{0.7}{R}\right) R$ being in kilohms; C in pF; and τ in nanoseconds.

CLEAR PULSE

The positive going' edge is differentiated by C2 and R2 (Fig. 18c), the negative going spike turning off TR1 whose output is a positive pulse situated in the middle of the sync period (Fig. 18d). This pulse



is applied to two parallel NAND gates whose negative going output (Fig. 18e) drives the "clear" inputs of the register.

The action of the "clear" pulse is to set the outputs (Q) of all six flip-flops to "0" awaiting the arrival of the first signal pulse which causes the output of the first flip-flop to go high (1) (since it takes an input from an inverted output (Q) for the period of the pulse and the interval period. The second signal pulse causes the second output to change to "1" and the first output to revert to "0".

Subsequent pulses appear individually as positive output pulses at the five flip-flop outputs.

The output from Q1 is shown at "B" in Fig. 19b.

SIXTH CHANNEL

The sixth channel output is obtained, if required, by the use of a further quadruple NAND gate wired as shown in Fig. 19a and if desired, associated with a servo unit. The positive going input signal at "A" is gated with the output of the flip-flop output Q1 to produce the signal shown in "C". This is again gated with the input to produce a negative going pulse as at "D", which is inverted, using another gate to produce the sixth positive going signal output at "E".

The shift register is formed by connecting the dual flip-flops in cascade as shown in Fig. 17a, it being noted that one interconnection pair is reversed to store the necessary "1".

Fig. 20 shows the printed circuit layout, the components being assembled as shown in Fig. 21. Note again the orientation of the integrated circuits and polarity of the capacitors and diode. When carefully soldered, thin insulated wire should be used to wire up the signal connection pads, the positive supply connections and the earth connections (Fig. 22).

DECODER VARIATIONS

The shift register can be extended for as many outputs (channels) as required provided that the "fan out" performance of parallel gates exceeds the number of loads—two per "clock" and "clear" input.

The present cost of SN7473N i.c.s invites the use of a cheaper gate to obtain the last output (Fig. 19) but a single flip-flop (FJJ111) could be used to store the "1".

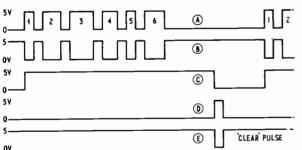
If Coder 2B (with positive sync pulse) is used the circuit in Fig. 17b is modified and the inverted (\overline{Q}) output from pin 6 of IC5 is applied to the base of TR1 and an output is taken from its collector (collector load 5.1kΩ) direct to the "clear" inputs.

The "clear" input signal (which overrides the action of "clock" pulses) is maintained to the end of the cycle.

A 1,000pF capacitor should be connected between the clock inputs and ground (0V) to prevent spurious ringing and the "extra" flip-flop method should be employed.

CONSTRUCTION NOTES

It cannot be over-emphasised that care should be taken with construction, particularly the soldering; if sound components are used and the circuit wired correctly, no difficulty may be expected; however, a simple solder bridge between copper pads can cause



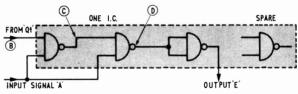


Fig. 19a. Quad 2-input NAND gate is used to obtain the sixth channel from the Q1 output. Waveforms are given below

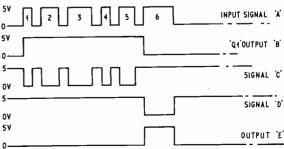


Fig. 18a. A typical input to first two NAND gates in IC1

Fig. 18b. Output of first two NAND gates fed into the clock line and into the signal period detector multi-vibrator IC5

Fig. 18c. Output Q from the resettable monostable multivibrator IC5 $\,$

Fig. 18d. Output waveform from TR1 indicating the positive pulse in the middle of the sync period

Fig. 18e. The pulse in Fig. 18d is inverted by two more paralleled NAND gates to produce a sharp "clear" pulse

Fig. 19b. Waveforms at input and outputs of the gates

complete malfunction, necessitating the use of an oscilloscope to locate the fault.

The following notes may be of value:

- 1. The i.c.s should be obtained from a reliable source and properly identified. It is difficult to remove a "dud" and the copper printing may be stripped with excess heat.
- 2. The capacitors should be inspected for leaking and loose ends and then should be polarised before assembly. A "working" voltage should be applied from a battery via a current meter noting that the leakage current falls to a low value: the absence of an initial charging current indicates an open circuit capacitor and an unpolarised or short circuit capacitor will cause excess current causing i.c. failure. Tantalum capacitors or long life electrolytics are preferred.
- 3. The wire ends of the diodes should be carefully bent with wiring pliers to avoid fracture of the glass seals.
- 4. Soldering should not be attempted without a clean small bit.
- 5. The recommended technique for insulating wire connections is to bare the ends of an excess length of multi-stranded wire, twist up and solder the ends, causing the insulation to creep back beyond the point where baring took place. The ends are then cropped till doin of clean soldered wire is exposed: the iron is applied to the tinned copper pad until it runs freely; then the wire is applied.

TESTING

The components should be inserted on the board and wired up as indicated, noting and checking the polarity of each component as fitted. Before any interconnecting wires are fitted the board should be held to the light and the component connections checked against the theoretical circuit. The variable resistors (coders) may be replaced by a selection of fixed resistors (as shown in photograph) within the VR range and the current supply voltage applied, noting the current to be less than 150mA. If an oscilloscope is available the waveform can now be examined at the output, otherwise an audio test may be applied (capacitor and headphones) noting the output of the multivibrator, each stage and at final output. The coder can now be connected to the decoder as in Fig. 1 noting that the supply current to the decoder is less than 50mA.

An oscilloscope triggered from the coder output should show the pulse train input and individual decoder output pulses.

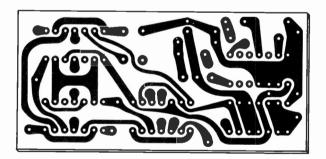
A d.c. meter (of 20 k Ω per volt resistance or better) should indicate different voltages of the order of $\frac{1}{2}$ to $\frac{3}{4}$ V at each decoder output in accordance with the values of coder resistance used. The meter performing an integrating action on the 5V pulses, being present, say, for 4 milliseconds every 30 or $4/30 \times 5$ V. The variable resistors may now be fitted and their action noted in changing the decoded pulse length or d.c. output.

Note: In Part 1 last month the following corrections are necessary:

Fig. 5. Capacitor C15 should be reversed

Fig. 7. The blob of ink at bottom centre is not part of the printed circuit pattern

 Table 1. The quad 2-input NAND gate by STC-ITT should be MIC9465D

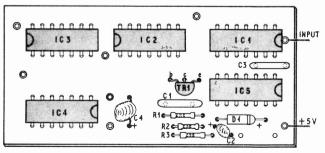


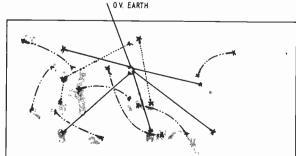
LOOK FOR THE FOLLOW-UP ARTICLE MODEL SERVO CONTROL IN NEXT MONTH'S ISSUE

Fig. 20 (left). Printed circuit pattern for the decoder

Fig. 21, Component layout of the decoder board

Fig. 22 (below). Link wires are soldered on to the copper side for power supplies and signal lines





Practical Electronics January 1972



A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought. This is YOUR page and any idea published will be awarded payment according to its merits.

STABILISED SUPPLY WITH OVERLOAD PROTECTION

The circuit diagram Fig. 1 was designed to power a 12 watt stereo amplifier. It was found to be desirable to have a constant voltage regardless of power being delivered by the amplifier. The circuit in Fig. 1 was evolved to achieve these requirements.

The potential at point A is always a fixed fraction of V_0 and is determined by R2 and R3. As R_1 , is lowered V_0 will lower causing the potential at A to decrease. As the base potential of TR1 is fixed by the Zener and VR1 so TR1's base-emitter diode D4 becomes more forward biased. Hence more current flows into the Darlington pair TR2 and TR3 which supplies R_1 with a greater current, restoring the original output voltage.

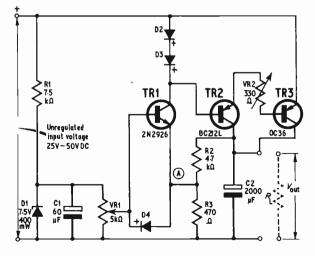


Fig. 1. Circuit diagram of the stabilised supply overload protection

Components VR2, D2 and D3 provide a current limit which is set by adjustment of VR2. Diodes D2 and D3 should be silicon types and have turn-on voltages of 0.6V each. Transistor TR3 is a germanium transistor and so V_{be} will be about 0.2V while \dot{V}_{be} for TR2 will be about 0.6V.

When a small current is being supplied to the load the voltage drop across D2 and D3 is not large enough to affect the circuit, but if R_L becomes very small the voltage drop across VR2 is arranged to bring D2 and D3 into conduction, preventing an excessive current through TR3. A sharper limiting action could be achieved by replacing D2 and D3 with a zener and increasing the value of VR2.

The output voltage is set to the required value by VR1. To obtain low voltages, say 10V, a bleeder resistor should be inserted across the output as TR3 has a leakage current of about 4mA. Capacitor C1 is included so that the output voltage rises slowly when the circuit is switched on. Diode D1 is included to protect TR1 when the circuit is switched off.

The output resistance of the circuit has been calculated and is given by:

$$R_{\rm o} = \frac{\rm R2 + R3}{\rm h_{FE(TR2)} \times \rm h_{FE(TR3)}}$$

This is 3Ω at the worst and about 0.5Ω at the best.

J. Welch, Northampton.

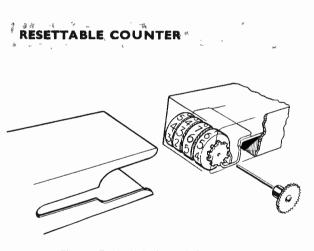


Fig. 1. Exploded view of the counter

HAVE converted an ordinary electromagnetic counter into a resettable counter by removing the spindle that supports the number wheels, and replacing it by one about $\frac{1}{2}$ in longer. On the protruding end I fixed a small knob. A slot was cut in the metal cover to enable it to be slid over the protruding spindle, see Fig. 1. The counter may now be reset by turning the knob.

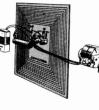
With some counters a second ratchet under the number wheels prevents them being reversed. A small lever may be fixed to the second ratchet to enable it to be disengaged when reversing the counter.

> A. Hartley, Bolton, Lancs.

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Audio Fair look at some recent developments in the hi fi industry By M. A. GOLWELL

A UDIO by definition means pertaining to sounds, so perhaps one could be forgiven for expecting noises resembling several kinds of phenomena. And since the Audio Fair has been described as an exhibition of home entertainment, perhaps we could likewise expect to see an assortment of musical instruments, television and radio, or—dare we suggest—electronic games. However, television was strictly banned by the organisers on the day of opening, so how far should we define "home entertainment" in terms of electronics?

Today's Audio Festival and Fair is a far cry from that high fidelity social function erstwhile held in luxury carpeted surroundings at the Hotel Russel and previously at the Waldorf.

LONDON OR BRIGHTON?

Because of overcrowding, the Audio Fair organisers were probably right in finding an alternative venue. Consequently in 1969 we heard some convincing arguments that Olympia *was* suitable; that acoustically treated demonstration rooms could be successful.

Give them their due, Rex Hassan and company tried hard amid tremendous criticism, but a great deal can be learned by closer examination of the Fair as a whole and its venue, and it is quite obvious that some more hard thinking must be done to rate this exhibition as near a success as can be achieved.

Now is the right time to take stock of such irritating discomforts as would not be tolerated in any hotel premises. Among these, I place catering at the head of the list, litter strewn gangways second, and poor ventilation third. And oh! that dreadful train service!

In the previous week I sampled the exhibition at Brighton called Inter/Nepcon and subsequently conclude that these premises must supersede Olympia by decades. Apart from the sweet ozone and spotless terraced houses nearby, the Metropole has every reason to be proud of its conference and exhibition centre and I recommend it wholeheartedly to organisers of such as the size of the Audio Fair.

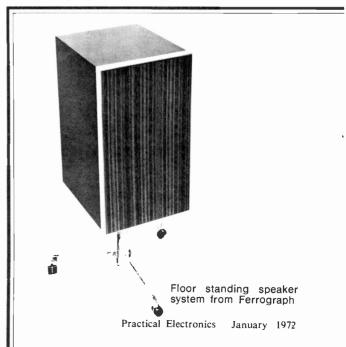
BROAD RANGE

The range of exhibits at this year's Fair has broadened to include medium fi, low fi, and even (if it is possible) no fi at all. The extension of range to include electro-musical gimmicks, record and tape shops, nameplates, and the not-so-high-as-it-used-tobe-fi inevitably lowers the general image that we have become so accustomed to in the past.

The overall technical standard of equipment and products shown was generally interfered with by infiltration of medium fi, but well established high quality products were still to be found.

RANK INFLUENCE

One interesting point to emerge is the influence of big business on the market through an established small company name. I am referring in particular



to the Rank Organisation whose acquisition of Wharfedale and Leak has already been geared towards mass marketing.

Although Harold Leak (now retired) did produce most of the parts that make up a complete hi fi system, he did not market a transcription unit. Logical first step then for Rank was to fill this gap, but not from their own stable. They have contracted with the Lenco Company of Switzerland to sell transcription units that they call the "Delta". Believe it or not, these are identical to the established Goldring Lenco GL75.

At least Rank recognises quality when it comes to such deals but it is a smack in the eye for the reputation of the former all-British Leak company. A Shure M75 magnetic cartridge is fitted to the Lenco pick-up arm, which has the usual cueing.

NEW SPEAKER SYSTEMS

Another first timer from Ferrograph, who previously preferred to specialise mainly in the high quality tape recording field, is a loudspeaker system. It is a three-unit assembly designed in collaboration with a university acoustics team, and uses two Goodmans treble units and one KEF B139 woofer.

Ferrograph state that their sophisticated crossover network eliminates what little coloration exists outside the pass bands. From this we can deduce that bandpass cut-off is very sharp so that interaction between speaker units is minimal. They quote a "sensibly flat" response from 30Hz to 15kHz.

Metrosound Audio Products Ltd. also announces a new speaker, the Duplex 25, to match their new ST60 amplifier (or any other make). With one conventional cone type unit for the low frequency end, they have incorporated a new electrostatic unit based on a new patent specification. High frequency dispersion is said to be improved by mounting this behind a lenticular grille.

The Duplex 25 is $27\frac{1}{2}$ in high and can handle 25 watts. Nominal impedance is 8 ohms and frequency range is quoted without reference level as 40Hz to 18kHz. Price at time of going to press £52.

HI FI IN KIT FORM

Looking for further outlets for their existing speaker kits, EMI Sound Products Ltd. are now marketing a range of enclosures in kit form from $\pounds 5.80$ for a compact bookshelf model to $\pounds 29.50$ for a large floor standing model. A hammer and screwdriver are the only tools needed to assemble these; the baffle-boards are already cut to suit the appropriate kits.

Heath (Gloucester) Ltd., who sell the famous Heathkits, announce a new tuner-amplifier kit, the AR-2000. The tuner processes a.m. (L.W., M.W., and S.W.) and f.m. mono and stereo broadcast signals, using up-to-date techniques, including f.e.t front end and integrated circuits for the f.m. i.f. and decoder stages. Ceramic filters are also used.

The kit price at £89.90 is competitive with equivalent commercial ready made units, but is still a little expensive for a kit.

Another well established name in hi fi circles is looking to expansion at a rapid rate. No less than eight new products were shown on the Goodmans stand.

Well known for its wide range of quality loudspeakers, they are competing forcefully with the likes of the Rank Organisation. The new items include an attractive a.m./f.m. tuner amplifier the Model One-Ten at £135; the Module 80 Compact, a combined f.m. stereo tuner amplifier of high performance with record player type GL75 at £165; the Dimension-8 bi-directional speaker at £69 teak, £75 white finish.

Goodmans also introduced a speaker kit, the DIN20 incorporating an 8in bass unit and an h.f. dome radiator with a crossover unit. This is a budget price kit of excellent quality for those with limited financial resources at £12.54 including purchase tax. Goodmans also expect to, release their new professional stereo tape recorder in 1972 at a price in the region of £400. The prototype shown at the Fair looks very impressive and uses slider controls, twin capstans, and switchable 2-speed NAB or DIN equalisation.

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	Metrosound ST60 amplifier	AT	
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Duplex 25 electrostatic and cone speaker from Metrosound	Heathkit tuner amplifier type AR-2000		
Practical Electronics January 1972		57	



BELT DRIVE

Past experiments on belt drive turn-tables have met with cynical comment, but progress to date is brought to fruition in the GL85 integrated transcription turntable unit and pick-up arm from Goldring.

The turntable is driven by a "non-stretch" flat belt from a low speed 16-pole synchronous motor. Motor speed is accurately controlled by an electronic stabiliser circuit. Two speeds are available and an illuminated stroboscope indicates speed accuracy.

An electronic sensing device raises the pick-up arm at the end of the record and switches off the motor. This unit is expected to be available in January and is to be supplied complete with plinth for \pm 72.78 including purchase tax.

The Connoisseur BD2 turntable at £32 now has push button speed change.

STEREO PHONES

Among headphones on show, which seem now to be a common occurrence for demonstration at the Fair, Koss were dominant with the excellent PRO-4AA stereo phones popping up on several stands, not the least being our own. The first four-channel headphones were also demonstrated by Tape Music Distributors Ltd. the distributors for Koss. They are the K2+2, using four separate dynamic drivers, and are likely to sell for around £45.

A.K.G. introduced the K180 (at £30) high quality headset with balancing volume controls.

Wilmex were demonstrating the Japanese Stax SR3 electrostratic headset with their power driver switch-over unit type SRD5. At 420 grams these are comparatively lightweight. The SRD5 will power two pairs of SR3 phones. A complete set of one pair of phones and one power unit costs £45:40; a second pair of phones would cost £31. In spite of reassurances I don't particularly care for high voltages so close to my ears.

DEMONSTRATIONS

The Hi-Fi Theatre is an excellent way of getting to know people and several personalities in the world of audio engineering, with a selection of magazine writers, gave some fascinating demonstrations, albeit, we are pleased to record, without commercial prejudice.

This is an essential alternative to an otherwise biased demonstration that may be set up on manufacturers' stands. The range was very wide and we were pleased to note the "free-for-all" session on the Friday where various magazines collaborated to help the public audience with their hi fi problems. This session was later edited and broadcast on BBC Radio London in November.

WE WERE THERE

What of PRACTICAL ELECTRONICS? Judging by the sales of P.E. Gemini Amplifier reprints, especially at the exhibition, this project is still a very popular do-it-yourself amplifier design.

The amplifier, originally shown last year, was joined this year by its new companion the P.E. Gemini Tuner, a stereo f.m. tuner of compatible design, with integrated circuit phase lock loop decoder and high selectivity ceramic filters.

Full details of this new design will appear in PRACTICAL ELECTRONICS very soon.

Also represented on the P.E. stand was our new companion EVERYDAY ELECTRONICS, which showed among other items, the integrated circuit record player, that appeared in the November issue.

Finally, a note for your new 1972 diary; the other London audio exhibition, Sonex at Skyway Hotel, near London Airport, will be open to the public March 24 to 26 inclusive, preceded by two "trade only" days.

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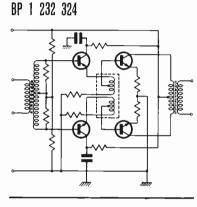
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IMPROVING AMPLIFIER PERFORMANCE

N their new British Patent 1 232 324, Rediton Limited discuss multistage transistor amplitiers of the push-pull type, capable of operating over a wide frequency range with coupling of the stages directly and by an inductive device with a view to improving performance.



Redifon explain that it is already known that a pair of transistors can be biased into class A conditions and coupled by way of a transformer so as to drive a further pair of transistors biased into class B conditions. It is also known that a pair of transistors can be biased into class B conditions and directly coupled to a further pair of transistors biased also into class B.

But, according to Redifon, in the transformer coupled arrangement there is difficulty in providing a driver transformer for the class B stage with a sufficiently low leakage reactance to transfer power over a wide frequency band with little distortion.

Various ways round problems like these have been suggested, but Redifon claim they have a novel idea in suggesting an amplifier with a first stage in class A and a second stage in class B, using an inductive device with a pair of tapped windings, arranged so that an output electrode of a transistor in the first branch of the first stage is connected to one end of one of the tapped windings, and an output electrode of a transistor of the second branch of the first stage is connected to one end of the other of these windings.

An input of a transistor in the first branch of the second stage is

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connected to the tapping of the first tapped winding and an input electrode of a transistor in the second branch of the second stage is connected to the tapping of the other one of the tapped windings.

The two windings of the inductive device are connected in a sense such that power for driving each transistor of the second stage is provided by both the transistors of the first stage.

Redifon give a circuit diagram for an amplifier of this type and the specification also carries some specific component values. Because of the load sharing by the transistors of the first stage, regulation is improved and distortion due to the effect of intermittent load imposed by each second stage transistor, is reduced.

SYNTHETIC ANIMAL SOUNDS

BP 1 228 405

N the name of Santa Rita Technology Inc. of California, BP 1 228 405 is particularly interesting because of its possible application to the projected airport at Foulness.

As is well known it is often necessary to rid airport runways of birds. This can be done by generating alarm signals, but, as Santa Rita explain, recording natural animal sounds and replaying them over a public address system is quite often only shortlived in its effect. The birds in some strange way seem to get used to the recording and recognise it as phoney.

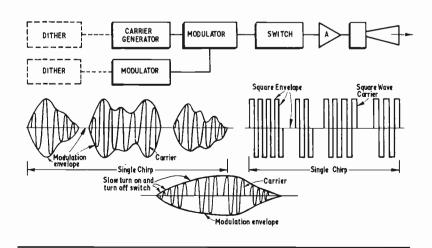
Their new invention is a system for producing synthetic animal sounds and in its simplest form it comprises a signal generator for generating carrier signals in the frequency range 500Hz to 5,000Hz. The selected carrier is amplitude modulated by another tone in the range 50Hz to 300Hz and a straightforward electro-acoustic transducer is used to convert the modulated carrier signal into a sound signal.

The sound generator can be arranged to turn the modulated carrier signal on and off at a rate of one to five times per second either sharply or gradually. And it may also have a dither arrangement for dithering the frequency of the modulating means over a two-to-one range. Moreover, the carrier signal can also be dithered over a frequency range comparable to the frequency of modulation.

Consequently, it is apparently easy to vary the sound signal components and so avoid the birds getting used to any one signal. The specification shows a block

The specification shows a block diagram of the basic layout. Some waveforms show how the "chirps" produced can be either in square wave or sinusoidal form.

Santa Rita claim special success in the diversion of red-winged blackbirds from a landing strip for jets with a 3,000Hz carrier and a 100Hz dithered random square wave amplitude modulation signal.





Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

RECHARGEABLE BATTERIES

Although rechargeable nickel cadmium cells are now quite common, it is claimed that the latest range, type NCC, from Ever Ready can be recharged to half their nominal capacity in 60 seconds using standard HP2 cells.

Now available to the constructor, the prices vary according to capacity and quantity ordered. Typically a U2 replacement would cost in the region of $\pounds 2$ for a single cell but would be proportionately reduced for larger quantities.

Button cells are normally supplied for higher voltage work and typically a 4.8V 550mAh cell, as used in radio control, would cost approximately £3.30.

Full technical details of the NCC range and stockists can be obtained from Ever Ready Company (Great Britain) Ltd, Special Battery Division, Hockley, Essex.

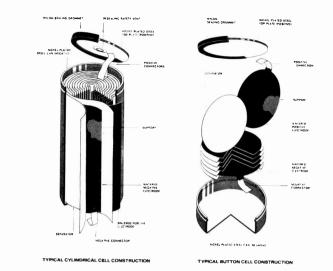
INVEST IN CATALOGUES

Time and again we get readers writing in asking where they can obtain certain items, ranging from resistors to infra-red emitting diodes. Apart from telling readers the source of supply we suggest that they should invest in some of the excellent components catalogues available from advertisers.

Typical of these are the latest component catalogues from such firms as: Electrovalue (64 pages, 10p), Home Radio (Components) Ltd (315 pages, 50p at shop or 70p by post), Henry's Radio Ltd. (350 pages, 40p at shop, 55p by post), LST Electronic Components Ltd (44 pages, no charge), and G. W. Smith and Co. (Radio) Ltd. (272 pages, 37½ p at shop, 47½ p by post).

The catalogues from Home Radio, Henry's and G. W. Smith contain special redeemable vouchers to help off-set the outlay for the catalogue.

The LST catalogue, available free, have added several new items to its opto-electronics section including inexpensive gallium arsenide light sources and planar silicon photo transistors.



Cell construction for the Ever Ready NCC range

There are many other interesting items listed in the LST catalogue including details of their data sheet service.

For our Midland readers, who often complain that they have to order from our London and Southern based advertisers, **Hawnt** & Co. Ltd. publish an excellent components catalogue.

Their latest edition (122 pages) contains sections on semiconductors, integrated circuits, capacitors, resistors and ferrites. Further details can be obtained from Hawnt & Co. Ltd., 112–114 Pritchett Street, Birmingham, B6 4EN.

VERSATILE PRINTED CIRCUIT BOARDS

The art of being able to pack a large number of circuit components on a small circuit board is a problem that plagues the professional as well as the amateur experimenter.

This problem has certainly been tackled, but whether conquered remains to be seen, by **NIP** Electronics with the introduction of a range of small printed circuit boards called "Nippiboards".

The board layout is based on the concept that transistor circuits follow common parameters for interconnection, with the addition of occasional requirements for feedback paths and cross-coupling. These paths are provided on the basic interconnection pattern, making it unnecessary to cut away the copper.

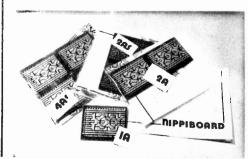
Link wires are kept to an absolute minimum and the board layout will, in most cases, follow closely the circuit diagram. This makes for easier circuit checking and servicing. There are two ranges available: range "A" has an s.r.b.p. base and range "B" which has a fibreglass base. The s.r.b.p. base is cheaper and easy to work with, but the fibreglass base overcomes some of the problems of extreme atmospheric conditions. All boards are flux varnished to keep the copper clean and make soldering easier. Range "A" costs from 15p: range "B" from 18p for one basic pattern.

Typical of the kind of circuits that can be built up on the boards are amplifiers, oscillators, tone circuits, switching circuits, timers, car and home electronic aids. One basic pattern will accommodate up to four transistor stages, whether they be *pnp. npn.*, f.e.t. or u.j.t. devices. Multiple pattern boards are also available for more complex circuits.

To date, the range of boards is only designed for transistor work. But we understand that a range designed specially for integrated circuits will be launched in the near future.

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N Part 3 the two aerial phase-switched interferometer was briefly described. It is now necessary to deal with this system in more detail since it is the next ste pin the project for those who have the space for it, as the same requirements as to base line apply here. The difference between the simple two-aerial system and the phase-switched system is the extra electronic units necessary and the aerial feeder modifications.

THE PHASE-SWITCHED INTERFEROMETER

When two aerials are spaced an exact number of wavelengths apart the aerial pattern will be such that the centre lobe will be at a maximum in the plane of symmetry. At this point the pattern is exactly the same as the simple interferometer, where the lengths of the feeders are exactly equal to each other. If now a length of cable equal to one half of a wavelength is introduced in one feeder line the pattern will change so that there is a displacement of the lobes. This displacement will bring the minimum point between lobes to the centre of the plane of symmetry, see Figs. 8.1a, b.

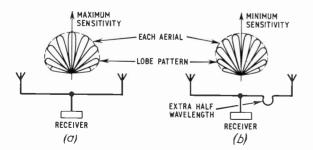


Fig. 8.1. The introduction of one half wavelength piece of cable into one feeder line displaces the aerial pattern as shown in (b)

As the aerial is moved with the rotation of the earth the aerial pattern can be changed at a predetermined rate so that if there is a source of radiation in the aerial beam it can be scanned. The alternate "putting in" and "taking out" of the half wave section which does the scanning can be done by using an electronic switching system.

PHASE SWITCHING SYSTEM

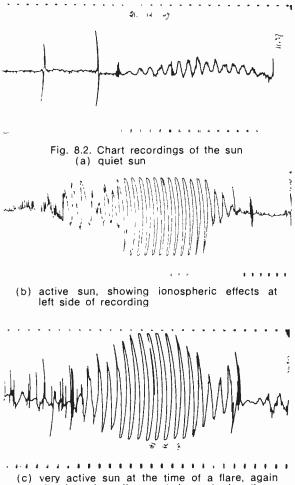
The number of times that the half wave section is switched in and out of circuit is largely a matter of choice, and to some extent depends on the purpose of the project under study. The circuits that will be given will have different speeds of switch operation, but the results on the recorder will not noticeably vary. The record that appears on the chart is an indication of the intensity of the radiation in terms of the modulation of that radiation at the switching frequency. If the switch frequency is, say, 900Hz, then the radiation received by the aerial will have a component at 900Hz superimposed upon it. This can be amplified and detected in the system and the result recorded on the charts as shown in Figs. 8.2a, b, c. The pattern that appears is the result of the sequence through which the input to the receiver goes before being recorded. This sequence is as follows:

With the source of radiation in the centre of a lobe, that is at maximum sensitivity, the length of the paths will be equal and the aerials will each send to the receiver the same level of signal. The two levels will add together so that the receiver will receive twice the value of the signal. When the half wave section is added the paths now being unequal, the resultant signal will be zero. Therefore the actual amplitude of the signal reverses in phase with the adding and subtracting of the half wave section.

The switching frequency is controlled by a circuit which provides a square wave to operate the switch and at the same time to operate a synchronous detector which reverses the phase of the output in unison with the switch. The block diagram Fig. 8.3 shows the individual units in which it will be seen that in addition to the units already in operation, there are four new ones.

The switch generator is a multivibrator of the usual type, and may be a part of the unit which includes the narrow band amplifier and the synchronous detector. The actual arrangement is left to the individual constructor.

Each unit will now be described separately, starting with the phase switch for the aerial. There are variations of this: one type being a hybrid junction used as a balanced network made up largely of coaxial cable, and a more simple diode switch unit. Both give similar results but the hybrid junction needs more skill to construct and has to be replaced as a whole when the frequency is changed.



showing the effects of ionospheric disturbance

HYBRID JUNCTION PHASE SWITCH

'. The hybrid junction phase switch consists of a network of coaxial cable with the switching diodes as part of the assembly, see Fig. 8.4. The action of this switch can be understood by following the diagram and remembering that a "short" looks like an open circuit when seen through a quarter wavelength of transmission line.

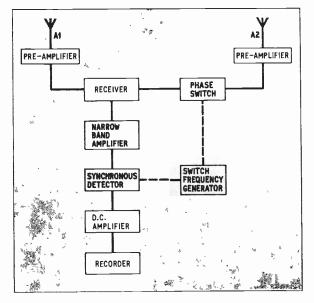


Fig. 8.3. Block diagram of the phase-switched interferometer. Units forming the original simple interferometer are shown in unshaded boxes

The switching frequency, which is provided by the switch frequency generator and its amplifier, is fed to the diodes via the choke and capacitor filter network. Suppose that the right-hand diode D2 is conducting, then at the junction of the quarter-wave section with the hybrid it will appear "open", the signal can therefore pass from the input via the first quarter-wave section through the three-quarter wave section to the output point and thence to the receiver.

At this time the left-hand diode D1 is non-conducting and therefore "open"; it appears as a short at the left-hand junction of the hybrid and no signal can pass. When the switch reverses then the left-hand side carries the signal and the right-hand side is blocked off.

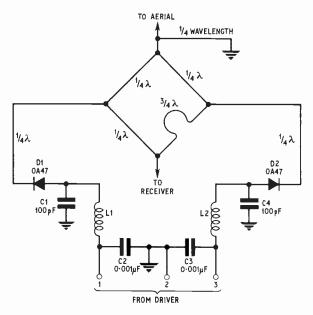


Fig. 8.4. A hybrid junction phase switch circuit



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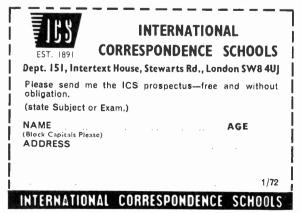
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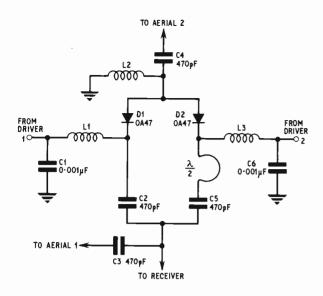


Fig. 8.5. A simple diode switch circuit

SIMPLE DIODE SWITCH

The simple diode switch consists of a network shown in Fig. 8.5 which uses diodes and a halfwavelength section for the frequency of operation. One advantage of this type of switch is that by having a number of half-wave sections, each for a different frequency, they can be conveniently plugged in to two aerial sockets thus making a very versatile assembly.

The sequence of operation in this switch is that the switching signal is fed into the points 1 and 2 via the filter network; in one direction a diode will conduct, say the left-hand side D1, at this time the other will be non-conducting so that the signal will pass via the conducting arm. When the switching voltage reverses, the diode on the right-hand side will conduct and the signal will pass via the halfwave section. When the switch is in operation the half-wave section will be alternately "in" and "out". The same conditions apply therefore as with the more complicated hybrid unit.

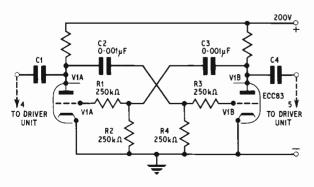


Fig. 8.6. Typical circuit for a switch frequency generator

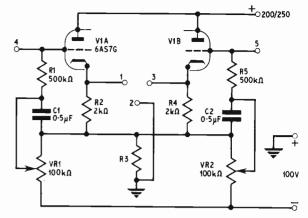


Fig. 8.7. Typical circuit for a driver amplifier for interposition between the switch frequency generator and the phase switch

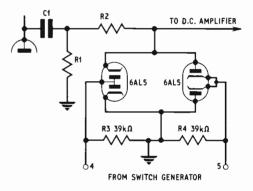


Fig. 8.8. Typical circuit for a phase sensistive detector

THE SWITCH FREQUENCY GENERATOR

The switch frequency generator is basically an audio oscillator operating at the frequency chosen for the switch. In this case it is a multivibrator which delivers about 30 volts peak-to-peak with a square wave form, and the circuit is shown in Fig. 8.6. The voltage output is sufficient for the operation of the synchronous detector, but needs a driver amplifier for the phase switch. The reason for this is that the diodes in the phase switch need to pass a current of the order of 50mA so that the power driver is necessary to accomplish this. The driver amplifier is shown in Fig. 8.7.

SYNCHRONOUS DETECTOR

As has been explained, the switching in and out of the half-wave section creates a condition where the receiver is fed with a signal which reverses in sign, hence the term phase switch. The reversal is in this case 180 degrees, so a system is required to follow this after the detection and amplification of the switch modulation. The diagram in Fig. 8.8

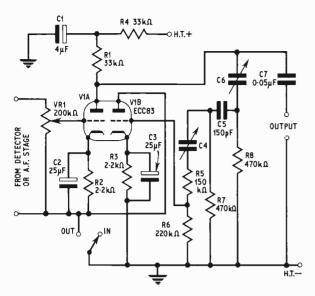


Fig. 8.9. Typical circuit for a narrow band amplifier set for switch frequency 900-1,000Hz

shows the details of the circuit for the synchronous detector, and this is fed from the narrow band amplifier and coupled direct to the d.c. amplifier.

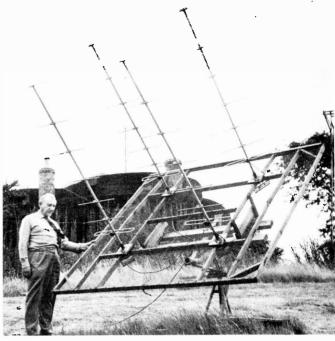
NARROW BAND AMPLIFIER

The narrow band amplifier is an important part of the system because it passes the signal at switch frequency and attenuates those signals which are not required. Strong signals from interference and radio signals which break through from commercial stations are largely eliminated.

The circuit shown in Fig. 8.9 is suitable for the switch frequency adopted. There may be a certain amount of insertion loss due to this amplifier; if this is so and the amplifier has been fed from the second detector of the receiver, then an additional stage of amplification can be interposed between the detector and the narrow band amplifier.

CONNECTING UP THE SYSTEM

The whole system is now ready to be connected up. It may be that there will be some snags in getting things right straight away. This raises the important point as to the stability of power supplies. If the units are checked individually for operation



A solar radio telescope aerial working at 200MHz

it should not be too difficult to get the system working satisfactorily. An oscilloscope would be a very useful adjunct in this connection, for all the systems can then be visually checked. For example the square wave pulses must be of equal width and amplitude. The phasing of the whole system can be more easily checked when an oscilloscope is used since each successive part of the system can be dealt with in sequence and any trouble quickly found.

Another point that perhaps should be emphasised again is the importance of good jointing and efficient bonding. Poor earth bonding can lead to many troubles, not least of which is spurious noise and breakthrough. If the area is one where there is a great deal of commercial activity it could be that there will be blocking of the pre-amplifiers. In this case it will be necessary to use some sort of filtering at the aerial input to offset this. Transistor preamplifiers are particularly prone to this trouble.

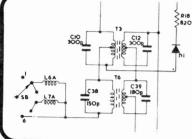
ACKNOWLEDGEMENT

The photographs of chart recordings Fig. 8.2 were taken by J. C. Codling, Clare Secondary School.

In the next article some more sophisticated units for the phase-switching system will be described.

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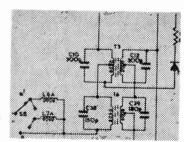


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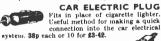
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OLFACTRONICS

I've just heard that a Nipponese company are making claim to an invention which amounts to an electronic "nose."

Naturally, smell sensors are not unknown, the Americans developed a device which could smell-out the Vietcong a few years back; this was a kind of funnel-shaped thing containing a microphone which had its fixing adjacent to a thin membrane. The membrane carried a number of sensitive (un-fed) bed-bugs which reacted to the natural body-odours of the enemy in a way which really had to be seen! The "tattoo" which they beat-out on the skin resulted in a signal that could be amplified after the microphone.

The device from Nippon, however, is quite different to previous sensors. For one thing, it can apparently differentiate between a number of different odours, and for another, it is entirely solid-state.

Several compounds are employed in the fabrication of the new device, typical of which are the oxides of in, zinc, and the sesquioxide of iron. Whenever such gases as hydrogen, carbon dioxide, acetylene, or propane are applied to the sensor its conductivity decreases sufficiently to give an easily noticed indication. often without further amplification!

We can only hope that the art of olfactronics (my handle) will soon become sufficiently sophisticated to counter the current wave of destruction by detecting the smell of gelignite before it blows someone apart.

COOL GLOW

Right "out of the blue" (well, probably nearer purple) someone has come up with a rather interesting way of making a draught. It all came about when Oscar Blomgren of Inter-Probe, Chicago, tried applying an electric field to the task of diverting an oxy-acetylene flame away from the inside of a blowtorch to prevent its deterioration. Instead, he got a cooling effect!

The cooling results from that funny effect you'll probably remember from experimenting with the school physics lab Wimshurst machine, electric wind. This, as you may recall, has nothing to do with flatulence but is associated with the purply-blue corona discharge that appears around any sharp, or pointed, parts of the machine's electrodes.

The vortex columns of air produced by the discharge are employed by Blomgren to cool the air near any heated surfaces. So efficient is this method of cooling that a glowdischarge produced by a 30kVpotential, at only 200μ A, is reported to have already reduced the temperature of an object from 1.675 degrees Fahrenheit to a mere 970 degrees Fahrenheit in a couple of seconds! An ideal device, one would think, where there was a special need for absence of motordriven fans with their attendant vibration. Such noiseless cool must be a real "blow" for fan-makers!



JABBER-JAMMER

It's quite on the cards that you can tolerate the majority of ambient noises like the intermittent tip-tap of the typewriter I am using right now, and the noise from a neighbour's transistor radio. Even so, the chances are that you will experience much more difficulty when attempting to concentrate on some abstruse problem when it is accompanied by the chatter of people in conversation nearby. How to overcome it is a perfectly reasonable question.

Just yesterday, I came to learn the answer to this, seemingly, difficult poser and Fig. I shows the easiest solution. The circuit comprises a simple audio oscillator connected direct to a hearing aid type earphone. The frequency of the device can be varied according to one's particular choice.

This done, it is then only necessary to bring the volume level up to a point which satisfies the compromise between being deafened by the oscillator, and going out of one's tiny mind as a result of the unwelcome intrusion into it by somebody's "yatter".

CHILD CARE

The burning need for something that will protect (in particular) the very young from the sometimes malevolent intentions of childsnatchers is a problem that has been in the limelight recently. Even when no harm is intended, the resulting anguish can often, and quite understandably, be unbearable.

From time to time, like others before me. I have made several vain efforts to solve this very real threat to youngsters. Among the ideas have been some which relied upon a weight-sensing switch in the child's pram, or a simple switch arrangement attached to the pram harness. Limitations, unfortunately, exist in these ideas since no facility was provided which sensed movement of the carriage. Perhaps another switch associated with the brake would be the answer; so it goes on, until, ultimately, one is tempted to give the problem up. But, there are quite a few other ways of tackling this enigma.

Not unnaturally, it would be well on the way to meeting our requirements if the abduction alarm had, in addition to warning ability, a fail-safe feature. It was in this vein that I chanced to come upon an idea that might just be the answer. If the child to be protected carried a low power radio control transmitter (complete with sewn-in antenna) in its coat, a signal would be transmitted, certainly to within the immediate locale of an interested parent.

It could be arranged that "mother" held, at home, a receiver tuned to the signal transmitted from the child and which operated a bell every time the transmitter went out of range. It would then be a simple matter to determine whenever junior was either in the course of being kidnapped, or had just plain toddled off somewhere. Who knows, it might just be better than sending up distress rockets!

Unfortunately, the Post Office is hardly likely to approve of this particular form of "model control" -Ed.



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02 03 04	221p 40251 221p 40309 321p 40310	221p BC169B 871p BC169C 321p BC170 45p BC171	119 BPX29 129 BPY10 129 BRY39 159 BSW41	\$1.80 NKT224 \$1.45 NKT225 47 p NKT229 42 p NKT237	200 TI853 2210 TI860 300 TI861 350 TI862	221p Car 221p i w 25p i w 271p i w	bon Film att 5%, 1p. att 5%, 1p. att 5%, 2p.	↓W & ↓W are E24 series. ↓W, 1W, & 2W	THERMI		
	Matc cking 12 p per or	der. Europe 25p. Prices subject to	alteration without	1219 extra per pa ir) Letter 65p (Mi prior notice.	sir. n.). Parcel £1.69 (№	(in.).	M/O 2% 4p. att 10%, 2ip. att 10%, 6p.	are E12 series.] watt 2%, 4p .	VA3705	871p mistors also in	stoci
01	-452 0161/ lex 21492	9/3 A. P		ALL & S		I.C. and !	Semiconducto	Comprehensive or price list (24 pa	_{ges)} MON 2 Thur 9	-FRI 9-5.3 9-1 SAT	

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Super IC-12

High fidelity Monolithic Integrated Circuit Amplifier

177

Two years ago Sinclair Radionics announced the World's first monolithic integrated circuit Hi-Fi amplifier, the IC.10. Now we are delighted to be able to introduce its successor, the Super IC.12. This 22 transistor unit has all the virtues of the original IC.10 plus the following advantages:

- 1. Higherpower.
- 2. Fewer external components.
- 3. Lower quiescent consumption.
- 4. Compatible with Project 60 modules
- 5. Specially designed built-in heat sink. No other heat sink needed.
- 6. Full output into 3, 4, 5 or 8 ohms.
- Works on any voltage from 6 to 28 volts without adjustment.
- 8. NEW 22 transistor circuit.

SINCLAIR GENERAL GUARANTEE Should you not be completely satisfield with your purchase when you receive it from us, return the goods without delay and your money will be refunded in full, including cost of return postage, at once and without question. Full service facilities are available to all Sinclair customers.

- Output power 6 watts RMS continuous (12 watts peak).
- Frequency Response 5 Hz to 100KHz \pm 1 dB.
- Total Harmonic Distortion Less than 1%. (Typical 0.1%) at all output powers and all frequencies in the audio band.
- Load Impedance 3 to 15 ohms.
- Power Gain 90dB (1,000,000,000 times) after feedback.
- Supply Voltage 6 to 28 volts (Sinclair PZ-5 or PZ-6 power supplies ideal).
- Size 22 x 45 x 28 mm including pins and heatsink.
- Input Impedance 250 Kohms nominal.

Quiescent current 8mA at 28 volts.

With the addition of only a very few external resistors and capacitors the Super IC.12 makes a complete high fidelity audio amplifier suitable for use with pick-up, F.M. tuner etc. Alternatively, for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC.12 ideal for battery operation.



Price, inc, FREE printed circuit board for mounting. **£2.98** Post free

Sinclair Radionics Ltd, London Rd, St, Ives Huntingdonshire PE17 4HJ Telephone St Ives (048 06) 4311



Sinclair Project 60

The World's leading range of high fidelity modules





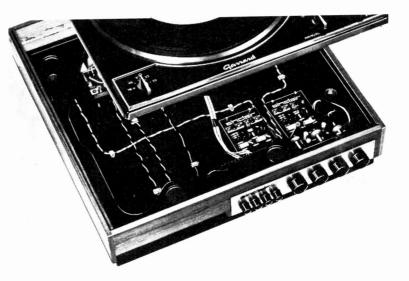


Complete Project 605 pack with comprehensive manual, post free £29.95 All you need for a superb 30 watt high fidelity

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stereo amplifier.





Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

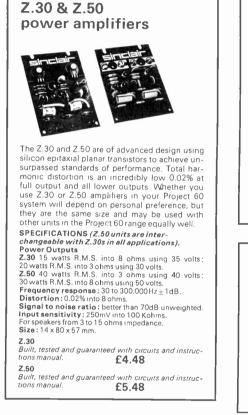
Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

Project 60 modules are more versatile – using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all – price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are finatstically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

Typical Project 60 applications

The Units to use	together with	Cost of Units
Z.30	Crystal P.U., 12V battery volume control	£4.48
Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U , F M, Tuner, Tape Deck, etc.	£26.90
2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Z.50, PZ.8, mains transformer	Mic, guitar, speakers, etc., controls	£19.43
	Z.30 Z.30, PZ.5 2 x Z.30s, Stereo 60, PZ.5 2 x Z.30s, Stereo 60, PZ.6 2 x Z.50s, Stereo 60, PZ.8, mains trsfrmr Z.50, PZ.8, mains transformer	Z.30 Crystal P.U., 12V battery volume control Z.30, PZ.5 Crystal or ceramic P.U. volume control etc. 2 x Z.30s, Stereo 60, PZ.5 P.U., F.M. Tuner, etc. 2 x Z.30s, Stereo 60, PZ.6 High quality ceramic or magnetic P.U., F.M. Tuner, etc. 2 x Z.50s, Stereo 60 High quality ceramic or magnetic P.U., F.M. Tuner, etc. 2 x Z.50s, Stereo 60 As above Z x Z.50, PZ.8, mains Mic., guitar, speakers, etc.

from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules





The Sinclair Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any const to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air+mail charged at cost

Project 60 Stereo F.M. Tuner



First in the world to use the phase lock loop principle

The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Good reception is possible in difficult areas, and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5dB. Sensitivity: 2μ V for 30dB quieting. 7μ V for lock-in over full deviation. Squelch level: 20μ V. A.F.C. range: \pm 200 KHz. Signal to noise ratio: > 65dB. Audio fre-quency response: 10 Hz = 15 KHz (\pm 1dB). Total harmonic distortion: 0.15% for 30% modula-tion. Stereo decoder operating level: 2μ V. Cross talk: 40dB. Output voltage: 2 x 150mV R.M.S. Operating voltage: 25-30 VDC Indicators: Power on/tuning/stereo. Size: 93 x 40 x 207 mm.

Built and tested. Post free.

£25

Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs. SPECIFICATIONS-Input sensitivities: Radio - up to 3mV, Mag. p.u. 3mV; correct to R.I.A.A SPECIFICATIONS—Input sensitivities: Radio - up to 3mV, Mag. p.u. 3mV: correct to R.I.A.A. curve ±1d8:20 to 25,000 Hz. Ceramic p.u. - up to 3mV: Aux - up to 3mV. Output: 250mV. Signal to noise ratio: better than 70d8. Channel matching: within 1d8. Tone controls: TREBLE + 15 to --15d8 at 10 KHz: BASS + 15 to --15d8 at 100Hz. Front panel: brushed aluminium with black knobs and controls. Size: 66x 40x 207mm. £9.98 Built tested and guaranteed.

A.F.U. High & Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s, and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less

loss of the wanted signal than has previously been possible. Amplitude and phase dis-tortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. - 3mA. H F. cut-off (-3dB) variable from 28KHz to 5KHz, L.F. cut-off (-3dB) Current variable from 25Hz to 100Hz. Distortion at 1KHz (35V, supply (0.02% at rated Built tested and guaranteed. £5.98 output. Size: 66 x 40 x 90 mm

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Please send	Name
	Address
I enclose cash/cheque/money order.	

Sinclair Q16/Micromatic

Q16 High fidelity loudspeaker

The Q16 employs the well proven acoustic principles specially developed by Sinclair in which a special driver assembly is meticulously matched to the characteristics of the uniquely designed cabinet. In reviewing this exclusive Sinclair design, technical journals have justly compared the O16 with much more expensive loudspeakers. Its shape enables the Q16 to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures. A solid teak surround with a special all-over cellular foam front is used as much for appearance as its ability to pass all audio frequencies without loss.

This elegantly designed shelf mounting speaker brings genuine high fidelity within reach of every music lover.

Specifications:

Construction: Special sealed seamless sound or pressure chamber with internal baffle.

Loading: up to 14 watts RMS.

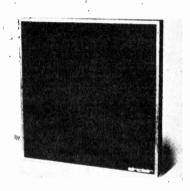
Input Impedance: 8 ohms.

Frequency response: From 60 to 16,000 Hz. confirmed by independently plotted B and K curve.

Driver unit: Special high compliance unit having massive ceramic magnet of 11,000 gauss, aluminium speech coil and special cone suspension for excellent transient response.

Size and styling: 9⁴/₂ in. square on face x 4⁴/₄ in, deep with neat pedestal base. Black all over cellular foam front with natural solid teak surround.

Price £8.98.



Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage AM receiver brilliantly designed to provide remarkable standards of selectivity, power and quality for its size. Powerful AGC counteracts fading from distant stations; bandspread at higher frequencies makes reception of Radio 1 easy. The plug-in magnetic earpiece provided, matches the Micromatic's output to give wonderful standards of reproduction. Everything including the special ferrite rod aerial and batteries is contained within the minute attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built and tested, you will find it as easy to take with you as your wrist watch, and dependable under the severest listening conditions.

Specifications:

Size: 36 x 33 x 13 mm (1.8 x 1.3 x 0.5 in.) **Weight:** including batteries, 28.4 gm (1 oz.)

Case: Black plastic with anodised aluminium front panel and spun aluminium dial.

Tuning: medium wave band with bandspread at higher frequencies (550 to 1.600 KHz).

Earpiece: Magnetic type.

On/off switching: By inserting and withdrawing earpiece plug.

Kit in pack with earpiece, case, instructions and solder **£2.48**.

Ready built, tested and guaranteed, with earpiece £2.98.

Two Mallory Mercury batteries type RM675 required from radio shops, chemists, etc.

To: SINCLAIR RADIONICS LTD LONDON	ROAD ST. IVES HUNTINGDONSHIRE PE17 4HJ
Please send	Name
	Address
for which I enclose cash/cheque/money order	P.E.1/72



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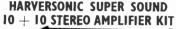




**Redeemable upon receipt of subsequent order value £5 or more, received before April 30, 1972









 STRUCTION

 A really first-class Hi-Fi Stereo Amplifier Kit. Uses 14 translators including Silicon Transistors in the first five stages on each channel resulting in even jower noise level with improved sensitivity. Integrated pre-amp with Bass, Treble and two Volume Controls. Suitable for use with Ceramic or Crystal cartridges. Output stage for any speakers from 5 to 15 ohms. Compact design, all parts supplied including drilled metal work, high parts supplied including drilled metal work, high epselfaction: Power output 14W r.ms. Per channel into 5 ohms. Frequency response ±3dB 12-30.000Hz. Bensitivity beter than 80mV into 1M G. Pullpower band-width ±3dB 12-15.000Hz. Bass boost approx, to * 12dB. Fuely detailed 7-page construction manual and parts list free with kit or send 18p plus large 8.A.E. FRUES AMPLIFIER KIT. £1050 P. & P. 15p. POWER PACK KIT. £3 P. & P. 30p. CABINET, £4 P. & 50p. CABINET, £5 P. & 50p. CAB

SPECIAL PURCHASE OF MANUFACTURER'S SUR-



SECIAL PURCHASE OF MANUFACTUREE'S SUR-PLUSH All Transistor F.M. tuner head with twin A.M. Gang incorporated. Beautifully engineered with precision geared reduction drive. F.M. R.F. Transistor, oscillator/Mixer and first L.F. stage (10.7 Mc/s output) with optional AFC connection. Built on printed circuit panel and fully acreened. Extremely stable over range 88-108 Mc/s. Brand new and pre-aligned. Nize 2(in H. : 14 in W. × 2(in D. For 6V D.C. %) 2.Mm.A. A.M. Gang fittel with trim-mers which can be connected to circuits if required. LiMITED NUMBER. Only 22.25 post free. Connection details supplied.

TELESCOPIC AERIALS WITH SWIVEL JOINT. Can b Brass. Extends from 6in. to approx. 221n. Maximum diameter in. 25p each. P. & P. 5p.

BRAND NEW MULTI-RATIO MAINS TRANSFORMERS. Giving 13 alternatives. Primary: 0-210-240V. Secon-dary combinations: 0-5-10-15-20-25-30-35-40-60V half wave at 1 amp or 10-0-10, 20-0-20, 30-0-30V, at 2 ampa full wave. Slze 3inL ~ 31inW × 3inD. Price \$1.75. P. & P. 30p

MAINS TRANSFORMER. For transistor power supplies. Pri. 200/240V. Sec. 9-0-9 at 500mA. 70p. P. & P. 13p. Pri. 200/240V. Sec. 12-0-12 at 1 amp. 83p. P. & P. 13p. Pri. 200/240V. Sec. 12-0-16 at 2 amp. 8138. P. & P. 30p. Tapped Primary 200-220-240V. Sec. 21:5V at 500mA. 68p. P. & P. 13p

4 AMP BATTERY CHARGER TRANSFORMER Brand new. For 6 or 12¹, 240V Prinary. Secondary volts r.m.s. off load 10 and 16-5V. Overall size approx. 21in × 21in × 31n. Weight 31b. Limited number at \$1-55. P. & P. 35p.

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HI-FI LOUDSTEARER SISTEM Beautifully made teak finish enclosure with most attractive Trgan-Vynair front. Size 16jin high ×10jin wide×647 deep. Fitted with E.M.T. Ceramic Magnet 13in×8in bass unit, two H.F. tweeter units and crossover. Power handling 10W. Available 3, 8 or 15 ohm impedance. Our Price £8:40 Carr. 65p.

CABINET AVAILABLE SEPARATELY $\underline{4450}$. Carr. 60p. Also available in 8 ohm with EMI I3in \times 8in. bass speaker with parasitic tweeter. $\underline{4050}$. Carr. 65p.

LOUDSPEAKER BARGAINS

LOUDSPEARER BARGAINS 5in 3 ohm 809, P. & P. 151, 7 × 4in 3 ohm 81.05, P. & P. 20p. 10 × 6in 3 or 15 ohm 81.90, P. & P. 30p. E.M. 8×5in 3 ohm with high flux magnet 81.62, P. A P. 20p. E.M. 132, X Sin 3 ohm with high flux ceranic magnet 8210 (15 ohm 42.92), P. & P. 30p. E.M. 13×88 in, 3 or 8 or 15 ohm with two inbuilt tweeters and crossover net-work \$42.0, P. & P. 30p. E.M. 13×78 twin cone (parastatic tweeter) 8 ohm \$2.25, P. & P. 30p.

BRAND NEW. 12th 15th H/D Speakers, 3 or 15 ohm. Current production by well-known British maker. Now with Hifux ceranic ferrobar magnet assembly **25-50**. Guitar models: 25th **26 50**, 35th **28**, **50**, 7. & P. 38th peach. E.M.I. 34 in HEAVY DUTY TWEETERS. Powerful ceramic magnet. Available in 3, 8 or 15 ohm 98p each. Powerful ceramic magnet. P. & P. 13p.

12in "RA" TWIN CONE LOUDSPEAKER 10 watts peak handling, 3, 8 or 15 ohn, 52, 20, P. & P. 30p. 35 ohm SPEAKERS 3'. ONLY 63p. P. & P. 13p. VYNAIR & REXINE SPEAKERS & CABINET FABRICS 54 in in. wide. Usually £1:75 yd., our price 75p yd. P. & P. 15p (min. 1 yd.). S.A.E. for samples. ngth.

SPEAKER CABINET BARGAIN SPEAKER CABINET BARGAIN Beautifully made all woolen construction medium walnut front, gold anodised expanded aluminium grille and dark sides. Approx. size overall 114 in high x Jin deep x 134 in wide at base. Baffle cut out for Sin speaker. Could accommodate amplifier or radio together with speaker. FEW LEFT ONLY \$1-38 post free.

HI-FI STEREO HEADPHONES

Adjustable headband with confortable flexifoan ear-muffs. Wired and fitted with standard stereo in jack plug. Frequency response 30-15,000Hz. Matching impedance 8-16 ohms. Easily converted for mono. PRICE 12.95, P. & P. 15p.

SINGLE HEADPHONE. With aluminium headband. Approx. 200 ohm. 25p. P. & P. 8p.

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GENERAL PURPOSE HIGH STABILITY TRAN-SISTOR PRE-AMPLIFIER. For P.U. Tape, Mike, Guitar, etc., and suitable for use with valve or transistor equipment. 9-18V. Battery or from H.T. line 200/300V. Frequency response 15Hz-25KHz. Gain 26dB. Solid encapsulation size 1½×1½×21n. Brand new — complete with instructions. Price 850. P. A. P. 130. nd new — con P. & P. 13p. 88p.

BRAND NEW E.M.I. LIGHTWEIGHT PICK-UP ARM WITH ARM REST. Fitted mono t/o stylus and cartridge for LP/78, ONLY 21. P. & P. Sp.

QUALTY RECORD PLAYER AMPLIFIER ME II A top-quality record player anaplifier employing heavy duty double wound mains transformer, ECC83, ELS4, and rectifier. Separate Bass, Treble and Volume controls. Complete with output transformer matched for 3 ohm speaker. Size 7in. w. 3 d. 6h. Ready built and tested. PRICE **43**75. P. & P. 40p. ALSO AVAILABLE mounted on board with output transformer and speaker ready to fit cabinet below. PRICE **54**88. P. & P. 50p.

DE LUXE QUALITY PORTABLE R/P CABINET MK II Uncut motor board mize 14; 121n., clearance 2in. below. bjio. bobwe will take above amplifher and any B.S.R. or GARARD changer or Single Player (except AT60 and SP25). Size 18×15×81m. PRICE 4475, P. & P. 509.



A styliably finished monaural amplifier with an output of 14 watts from 2 ELS4s in push-pull. Super reproduction of both music and speech, with negli-gible hum. Separate inputs for mike and gram allow records and announcements and announcements to follow each other



Fully shrouded section wound output transformer to Fully abrouded section wound output transformer to match 3-152 speaker and 2 independent volume controls, and separate base and treble controls are provided giving good lift and cut. Valve line-up 2 ELEA, ECCS3, EF86 and EZ80 rectifier. Simple instruction booklet 139 (Free with parts). All parts sold separately. ONLY 3*39, F, F, F, 550, Also available ready built and tested complete with std. input sockets, 539 F, F, F, F, S50.

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former matched for 3 ohm speaker. Separate volume control and now with improved wide range tone controls giving bass and treble lift and cut. Negative feedback line. Output 44 watts. Front panel can be detached and leads extended for remote mounting of controls. Complete with knobs, valves, etc., wired and tested for only \$475. P. & P. 355.





PARTASTIC "POLY PLANAR" WAFER-TYPE, WIDE RANCE ELECTRO-DYNAMIC SPEAKER Size only 114 in x 144 in x 14 in deep. Weight only 1902. Can replace cone speakers in any application and is particularly useful for those with limited space. Extremely rugged and shockproof. Operating temperature -20°F to +175°F. Power handling 20W r.m.s. (40W peak). Impedance & ohm only. Response 40H2-20kHz. Can be mounted on ceilings, walls, doors, under tables, for full details. Only 55-75 each. P. & P.25p.

A.c. mains 200-240 volts. Using heavy duty fully isolated

mains trans

former with full wave rectification

giving ade-quate snoothing with negligible hum.

line up:-2× ode Pentodes

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ECL86 1×EZ80 as rectifier. Two dual potentiometers are provided for bass and treble control, giving hass and

provided for bass and treble control, giving hass and treble boost and cut. A dual volume control is used. Balance of the left and right hand channels can be adjusted by means of a separate "balance" control fitted at the rear of the chasses. Input semitivity is separatively a separate "balance" control fitted (8 watts mono), into 3 ohr speakers. Full negative feedback in a carefully calculated circuit, allows high volume levels to be used with negligible distortion. Supplied complete with knobs, chassis size 111n. w.×41n. x. Overall height including valves 5in. Ready built and tested to a high standard. Price \$8.92. P. & P. 45p.

4-SPEED RECORD PLAYER BARGAINS Mains models. All brand new in maker's packing. LATEST BS.R. C109/A21 4-SPEED AUTOCHARGER. With latest mone compatible cartridge 56-97. Carr. 50p. With stere cartridge 57-97. Carr. 50p. SUITABLE PLIMTH UNIT FOR .BOVE with rigid plastic cart at the plimth UNIT FOR .BOVE with rigid plastic

SUITABLE FLINTH UNIT FOR ABOVE with rigid plastic cover. 55-75 complete. P. & P. 50p. LATEST GAREARD MODELS. All types available 1025. 2025, 5F25, 3000, A760 etc. S.A.E. for Latest Prices ! PLINTH UNITS cut out for (iarrard Models 1025, 2025, 2000, 3000, 3500, etc. With rigid transparent plastic cover. Special design enables above models to be used with cover in position. Also suitable for housing AT60 and SP25. OUR PRICE 55-75 complete. P. & P. 50p.

LATEST ACOS GP91/18C Mono Compatible Cartridge with t/o stylus for LP/EP/78. Universal mounting bracket.

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(Please write clearly) PLEASE NOTE: P. & P. CHARGES QUOTED APPLY TO U.K. ONLY. P. & P. ON OVERSEAS ORDERS CHARGED EXTRA.



3-VALVE AUDIO AMPLIFIER HA34 ME II Designed for Hi-Fi reproduc-tion of records. A.C. Maine operation. Ready built on plated heavy gauge metal chasis, size 74 in w.2 wind. X. 48 in. b. Incorporates ECC83. EL84. E280 valves. Heavy duty, double wound mains transformer and output trans-former matched for 3 obm-control and now with improved 111 Õ 0.

/78 mono or P. & P. 10p.

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R.C.S. STABILISED POWER PACK KITS



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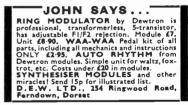
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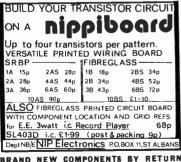
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