

## **ADCOLA** Soldering **Instruments** add to your efficiency

THE NEW 'INVADER'

**PRICE** 

£1.85

ADCOLA L.646 for Factory Bench Line Assembly

A precision instrument—supplied with standard 3/16" (4.75 mm) diameter, detachable copper chisel-face bit\*.

Standard temp. 360°c at 23

Special temps, from 250 c-410°c.

\*Additional Stock Bits

(illustrated) available

COPPER

B 38 1 - 3.2 mm CHISEL FACE B 14 3 - 2.4 mm B 24 1 - 4.75 mm SCREWORIVER B 12 14 - 4.75 mm EYELET BIT B 58 4 - 6.34 mm CHISEL FACE LONG LIFE B 42 LL 音 - 4 75 mm CHISEL FACE B 38 LL | - 3.2 mm CHISEL FACE B 14 LL 31 - 24 mm B 44 LL 3 - 4 75 mm SCREWORIVER

Don't take chances. We don't. All our ADCOLA Soldering Instruments are of impeccable quality. You can depend on ADCOLA day after day. That's why they're so popular. You get consistent good service . . . reliability . . . from our famous thermally controlled ADCOLA Element and the tough steel construction of this ideal production tool.



Write for price list and catalogue

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Safe, quick and secure it connects 2-core and 3-core bare-ended flexible leads to the mains (A.C. only).

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Safebloc saves time. No need to fit a plug for tests. No danger, as no current can pass with the lid open.

Invaluable for testing and demonstrations in industry and shops, the work bench and the home.

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BUILD 50 INTERESTING PROJECTS on a PRINTED CIRCUIT CHASSIS with PARTS and TRANSISTORS from your SPARES BOX

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Send a S.A.E. for full details and a brief description of all kits and Projects.

## LASHYS SCOOP

## HI-FI BARGAIN OF THE CENTURY rave up to 35% on LEAK recommended list prices!

As the United Kingdom's leading High-Fidelity Retailers, we are proud to offer you the "Scoop of the Century", Leak Hi-Fi Amplifiers, tuners, turntables and speakers at fantastic prices. Each piece of equipment is brand new and covered by Leak's full guarantee. Leak have been the U.K.'s leading Manufacturer since the birth of High-Fidelity—their quality has always been used as a yardstick for true High-Fidelity.



Leak Stereo 70 amplifier (cased only) Power output (both channels linewave driven): 35 watts r.m.s. each channel into 8 ohm loudspeakers. Total harmonic distortion: 0-1% for all power levels up to 25 watts r.m.s. each channel at IKHz into 8 ohm loudspeakers. Crosstalk: Between left and right channels 504B up to IKHz and -304B at 10KHz. Dimensions: 13in (W) 4½in (H) 9½in (D) LIST PRICE £75-00 LASKY'S PRICE £55-00 carriage and packing £1.00.



Leak Stereofetic FM tuner—Frequency range: 87 108mHz. Frequency response: \$\pm\$ 1dB 40Hz to 15kHz. Frequency drift: Less than 25KHz without A.F.C. Sensitivity: 2\frac{1}{2}\text{ microvolts for 3dB signal/noise}. Output impedance: 200 ohms. Distortion: Less than 0.5% for full deviation, i.e., less than \$\frac{1}{2}\text{/6}\text{ for average modulation}. Signal/Noise: 60dB A.M. suppression: 50dB. Dimensions (cased) 12in (W) \$\times 4\frac{1}{2}\text{in}\text{ (H) } \times 9\frac{1}{2}\text{in}\text{ (D)}\text{ (cased) } 1\frac{1}{2}\text{in}\text{ (H) } \times 7\frac{1}{2}\text{in}\text{ (D)}. CASED LIST PRICE 69:80 LASKY'S PRICE 645:00.

CHASSIS LIST PRICE £64:03 LASKY'S PRICE £39:50 carriage and packing £1.00 on chassis or cased model.



Leak Stereo 30 Plus amplifier (cased only) Power outputs (both channels sinewave driven) 15 watts r.m.s. each channel into 8 ohms loudspeakers. Music power outputs 20 watts r.m.s. each channel into 8 ohm loudspeakers. Total harmonic distortion 0.1% for all power levels up to 10 watts r.m.s. each channel at 1KHz into 8 ohm loudspeakers. Crosstalk: Between left and right channels 50d8 up to 1KHz and 30d8 at 10KHz. Dimensions: 13in (W) 4½in (H) 9½in (D). LIST PRICE £62-50 LASKY'S PRICE £45-00 carriage and packing £1,00.



Leak Truspeed Transcription Turntable Mk. III A low speed (250 r.p.m.) synchronous 12-pole hysteresis motor (100-130V, or 200-250V, 15mwA), gives constant turntable speed independent of mains voltage fluctuations. Speeds: 33½ r.p.m. and 45 r.p.m. Wow: Less than 0·15%, Flutter: Less than 0·02%, Frequency ranges: 20Hz-20KHz. Stereo separation: Better than 254B at IKHz. Dimensions: 12½in 15½in 7½in inc. cover. Complete with arm teak base and tinted perspex cover and Shure M75/6 cartridge. Made to sell for £69 50 LASKY'S PRICE £47-50 C. & P. £1-50.

Leak 206 speaker system—A Stainless Steel woven mesh grille cloth for maximum acoustic transparency and an attractive teak cabinet finished off with aluminium trim all round housing a 3 speaker system worthy of LEAK's name. Impedance 8 ohms, nominal. Frequency response: 60Hz to 18KHz. Power handling capacity 18 watts r.m.s., Resonance of bass unit (free air)'s 45Hz. Finish: Teak. Dimensions: 15½in 10in 8½in LIST PRICE £49:90 LASKY'S PRICE £31:50 (pair) C. & P. £2:00.



#### fantastic LEAK packages

	Package I
	Stereo 30 Plus (cased)         £62-50           Stereofetic (chassis)         £64-03
	Total Rec. Recail Price £126-53
ı	LASKY'S PRICE £84 C. & P. £1-50
	Package 2
ı	Stereo 70 (cased)         £75.00           Stereofetic (chassis)         £64.03
	Total Rec. Retail Price £139-03
	LASKY'S PRICE 494 C. & P. £1-50
ı	Package A
ı	LEAK \$T30 Plus (cased)
ı	LEAK TruspeedT/T system £69-50
i	Total Rec. Retail Price £181.90
ı	LASKY'S PRICE £120 post £2 00
	Package B
ı	LEAK ST70 (cased)
ł	LEAK Truspeed T/T system £69-50
ı	Total Rec. Retail Price £194-40 LASKY'S PRICE £130 C. & P. £2-00
۱	LASK 1'S PRICE 2130 C. & P. 22-00
ı	Package C
ł	LEAK ST30 Plus (cased)
ı	LEAK Stereoletic (cased)
ı	Total Rec. Retail Price £251-70
ı	LASKY'S PRICE £160 C. & P. £3-00
ı	Package D
ı	LEAK ST70 (cased)
ı	LEAK 200 5pks (pr)
ı	LEAK Truspeed T/T system 69-50 Total Rec. Retail Price (264-20
d	LACKVIC BRICE (170 C 0 D C) 00

ASKY'S PRICE £170 C. & P. £3-00

Package E	1.0
LEAK 5730 Plus (cased) Wharfedale Denton Spks. (pr). BSR McDonald MP60 Lasky's base and cover. AD76K magnetic cartridge Total Rec. Retail Price LASKY'S PRICE 192 C. & P. 67	£62:50 £39:90 £15:20 £4:75 £4:35 £126:70 2:00
Package F	
LEAK ST70 (cased) Wharfedale Triton Spks (pr) Garrard AP76 Garrard base and cover Shure M44/E magnetic cart Rec. Retail Price LASKY'S PRICE (129 C. & P.)	£75.00 £59.90 £27.85 £9.75 £11.63 £184.13
Package G LEAK ST30 Plus (cased) Lasky's Criterion Mk. X Spks. BSR McDonald MP60 Lasky's base and cover. AD76K magnetic cartridge Rec. Retail Price LASKY'S PRICE (85 C. & P. £)	£62-50 £25-00 £15-20 £4-75 £4-35 £111-80
Package H LEAK ST70 (cased) LEAK Stereofetic (cased) LEAK Truspeed T/T system LEAK 600 spks. (pr) Rec. Retail Price LASK YS PRICE (225 C. & P. &	£75.00 £69.80 £69.50 £99.00 £313.30
These are only a few of our age bargains incorporating	

equipment, many more can be arranged. If none of the packages here suit you please send us

details of your requirements and we will be pleased to quote.

# SUPER SCOOP!



DIGITAL CLOCK RADIO 8FC-59WA

SAVE £12-19

Sony's elegantly designed digital alarm clock radio. The digital clock shows you the time minute by minute with matchless accuracy, and once set the Digimatic will wake you up at the same time every morning without having to be re-set. The radio section features Sony's unique sleep button, you fall gently to sleep fulled by the sweet tone of the radio, which switches itself off at a predetermined time. Available in either black or white. Frequency range: 530-1,605KHz (AM): 87-108MHz (FM). Uses 8 transistors and 8 semi-conductors, built-in ferrite aerial and 3½ in loudspeaker. Power requirements: 230V a.c. 50Hz. Dimensions: 12½ in 3½ in 5½ in.

LIST PRICE LASKY'S **£16-75** Post £28.94 PRICE **£26.75** 

#### OUT NOW!'72 AUDIO-TRONICS



The great new 1972 edition of Lasky's famous Audio-Tronics Catalogue is now FREE on available request. The 44 newspaper size pages-many in full colour-are packed with 1,000's of items from the largest stocks in Great Britain of everything for the Radio and Hi-Fi enthusiast. Electronics hobbyist, Serviceman and Communications Ham. Over half the pages are devoted exclusively to every aspect of Hi-Fi (including Lasky's budget Stereo Systems and Package Tape recording Deals). and Audio accessories and don't miss LASKY'S AUDIO - TRONICS CREDIT CARD SCHEME offering holders one month's interest free credit up to £50 and the fantastic £1,000 plus colour TV competition. Send your name and address and 15p for post and inclusion on our regular mailing list.

## NEW from Carky

## CRITERION Mk. X

Because of the increasing demand for inexpensive demand for inexpensive, high quality bookshelf speakers, we have seen fit to introduce the Criterion Mk. X book-shelf system. The speaker a sealed infinite baffle type enclosure using 8in woofer, a 5in mid. range and 2½in tweeter. The compact tweeter. The com cabinet is finished oiled walnut and black woven speaker grille cloth. Frequency res-40Hz-20KHz. Power handling capacity: Max. 20 watts. Impedance: 8 ohms. The Criterion is fitted with two types of speaker lead connections screw terminal and phono. Size: 18¾in (H) × 9¾in (W) × 9¾in (D). Operational horizontally or vertically

LASKY'S AMAZING PRICE



£25

C & P 50p

#### a sound future

Lasky's Radio for over 38 years Great Britain's Leading Radio, High Fidelity, Tape Recorder and Electronics Specialists have vacancies in their West End and City Branches for both male and female Sales Assistants. We are seeking intelligent young men and women sales personnel to join our expanding organisation (already the largest in Europe) on a career basis with the finest prospects for early promotion and financial advancement. Working with our energetic sales teams in any of our six West End and City branches will bring you into contact with people from every walk of life including Pop Stars, film and television personalities, royalty and above all enthusiasts in every field of Audio, Hi-Fi and Electrohics-people who expect you to share their enthusiasm and interest.

Knowledge of Hi-Fi and electronics is not essential although it will help—all that we require as the basic qualities needed for a successful application are willingness to learn and a common-sense approach to the business of selling. Salary ranges from £850 to £1,850 plus, three weeks' holiday after one year, and Incentive Bonus Schemes are added benefits of working in our progressive organisation. Holiday arrangements will be honoured. If you are interested in a career in Audio and Hi-Fi write at once to: Kenneth Lasky, Lasky's Radio Limited, 3-15 Cavell Street, London El 2BN.

Interviews will be arranged to suit your convenience.

SUPER SPEAKER BARGAINS

WHARFEDALE DENTONS

A compact system sold in matched pairs for a perfectly balanced stereo system. Each Denton contains an 8in bass unit with 3in contains an 8in bass unit with 3in pressure unit, coupled by a Wharfedale crossover network. Rated input: 15 watts maximum. Frequencyresponse: 65-17,00Hz. Impedance: 4/8 ohms. Cabinet 9½in × 14in × 8½in. Available in oiled teak finish and are small enough to blend in with most surroundings.



LASKY'S PRICE £29.00 LIST PRICE

#### WHARFEDALE TRITONS

A new and exciting addition to the Wharfedale range, this three-speaker system will satisfy the most ardent Hi-Fi enthusiast. Shelf or floor standing—hand finished in oiled teak. Frequency response: 55-22,000Hz. C.A.B. Dome pressure unit. Bass unit Bin. Mid-range unit 5in. Treble unit 2 in. Impedance: 8 ohms. Size: 21½in × 9½in × 9in.

LIST PRICE £59-90

LASKY'S PRICE £39.00



#### **EXCLUSIVE TM-I**

MODEL TM-1 MINI-TESTER

The first of Lasky's new-look top value meters, the TM-lis a really tiny pocket multimeter providing "big" meter accuracy and performance. Precision movement calibrated to 3in of full scale. Click stop range selection switch. Beautifully designed and made impact resistant black case with white and metallic red/green figuring. Ohms zero adjustment.

- Size Only

  3½ in × 2½ in × 1½ in

  DC/V: 0-10-50-2501,000 at ik OPV

  AC/V: 0-10-50-2501,000 at ik OPV

  DC CURRENT: 0-1mA,

Resistance: 0-150k
 Decibels: -10dB to



ASKY'S PRICE £1.95 POST 13p

#### **5K ohms/V POCKET** MULTIMETER

Another new look pocket multimeter from Lasky's providing top quality and value. The "slimline" impact resistant case, size  $4\frac{1}{2}$  in  $\times 2\frac{1}{2}$  in  $\times 1\frac{1}{2}$  in fitted with extra large  $2\frac{1}{2}$  in square meter. Readability is superior on all low ranges: making this an excellent instrument for servicing transistorised equipment. Recessed click stop selection switch. Ohms zero adjustment. Buff finish with crystal clear meter cover.

- DC/V: 3-15-150-300-1,200 at 5k OPV AC/V: 6-30-300-600 at 2-5k/OPV Resistance: 0-10k/ohms; 0-1M/ohm Decibles: 10dB to 1-16dB Complete with test leads, battery
- AC/V: 6-30-300-600 at 2:5k/OPV
   DC Current 0-300μA, 0-300mA
  - and instructions

LASKY'S PRICE £2.95

#### LASKY'S NEW "LOW NOISE" CASSETTES FROM the U.S.A.

Model	Singles	5	10	20
C.60	32p	£1:52	€2.96	€5.60
C.90	50p	£2·37	£4·62	€8.75
C.120	69p	€3.58	£6∙38	£10.85

Post Each 5p. 5-20p. 10-25p. 20-50p.

#### TRIO KA.2002 PACKAGE DEALTOP VALUE

TRIO KA.2002 £39.50
Pair Wharfedale Dentons £39.90
BSR MP60 £15.20
LASKY BASE & COVER £4.75
AD76K CARTRIDGE £4.35
Total list price £103.70

#### PACKAGE PRICE £79.50

Carriage in U.K. £2:00
ADD £9 if Wharfedale TRITONS are preferred to Dentons

#### TMK MODEL 200 METER KIT

TMK offer the unique opportunity of building a really first-class precision multimeter at a worthwhile saving in cost. The cabinets are supplied with the meter scale and movement mounted in position. The highest quality components and 1% tolerance resistors are used throughout. Supplied complete with full constructional circuit and operating instructions.

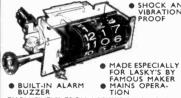
Specification

20,000 P.O.V. Multimeter. Features 24 measurement ranges with mirror scale accuracy. DCIV and current: 2%. Ac.W: 3%. Resistance 3%. Special 0-6V DC range for transistor circuit measurements.

LASKY'S PRICE £4.60

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#### DIGITAL CLOCK SCOOP



BUILT-IN ALARM BUZZER

SHOCK AND 12-HOUR ALARM
VIBRATION
PROOF

PROOF

HOURS, MINUTES
AND SECONDS
READ-OFF
FORWARD AND
BACKWARD
TIME ADJUSTMENT
LASKY'S BY

12-HOUR ALARM
AUTO-SCEEP'S
WITCH
HOURS, MINUTES
BACKWARD
TIME ADJUSTMENT
MENT

• SILENT OPERATION SYNCHRONOUS MOTOR

BUZZER TION

EXCLUSIVELY FROM LASKY'S in chassis form for you to mount in any housing. The clock measures 4½W X 1½H X 3½D (overall from front of drum to back of switch). SPEC.: 210/240V AC, 50Hz operation; switch rating 250V, 3A. Complete with instructions.

HUNDREDS OF APPLICATIONS. COMPLETE WITH KNOBS

ASKY'S PRICE £6.50

SPECIAL OUOTATIONS FOR QUANTITIES

#### TRIO HS.I HS.2. STEREO **HEADPHONE BARGAINS**

Models HS.1 and HS.2. Both these sets by TRIO offer really superb stereo reproduction in a lightweight, fully adjustable headset designed for optimum

adjustable headset designed for optimum comfort. Listening fatigue is unknown with TRIO headphones. Brief spec. both models: Input imp. 8 nominal (matching 4 to 8): max. input 0.5W: frequency response 20-19kHz: output sensitivity at 1mW input; HS.I 118BB, HS.2 111BB; weight 0.661lbs. Identical in appearance—both models are finished in ivory with contrasting foam-filled ear pads and head band. HS.1.

List Price £8-40. List Price £6-75.

Lasky's £5.00

Lasky's £4.00 Post IBP

Lasky's Radio Limited

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BAND (1:6-4mcs)

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transmitting and those planned for the future) both BBC and Commercial.

Hours and hours of endless enjoyment—a complete hobby in itself. One minute you can be listening to your favourite programme on BBC and then—at the flick of a switch—you can hear a crippled airliner being talked down to safety. YES! You can actually eavesdrop on the exciting conversations between pilot and control. MAKES A SMASHING CAR RADIO! Frequency range. Medium wave 540-1600kcs. FM-VHF 88-108m/cs, AIR-VHF 108-136m/cs. PUBLIC SERVICE BANDS-VHF 136-175m/cs, Automatic Frequency Control to pinpoint station—locks on and completely eliminates drift. 16 Transistors, 7 diodes, 1 thermistor, TWO AERIALS one internal ferrite rod and one external telescopic antenna. Uses standard batteries. Beautifully finished in black grained leatherette with the latest push-button internal territer od and one external telescopic antenna. Uses standard batteries. Beautifully finished in black grained leatherette with the latest push-button controls and slide rule tuning dials. Complete with hi-fidelity earphone and full written guarantee—spares always available. Size  $10 \text{ in} \times 6 \text{ in} \times 4 \text{ in}$ . Instant cash refund if not satisfied. Also available Battery/or mains Super de Luxe upright model.

Same frequencies—but with built-in mains adaptor and battery re-charger (no need to buy batteries again!)

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Brand new 44in wide × 16in deep × 18in high with legs. A superb piece of furniture. Carriage £1. WHILE STOCKS LAST.

CASH £35.50

#### HI-FI VALUE

(I) Garrard SP25 Mk. III: £II:50, P.P. 50p.

(2) Teak Plinth and Tinted Cover: £4-95, P.P. 35p.

(3) Sonatone 9TAHC Diamond Cartridge: £2.50.

1, 2, 3 Bargain Package, £17-95, P.P. 85p.



#### COMPONENTS MUST BE CLEARED

Transistor Radio Cases: 25p each. Size 9½in × 6½in × 3½in, Post 15p. Speakers: 35p. 2½in 8Ω. Brand new. Post 15p.

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Pots: 25p each. Post 5p. D/SW 500/500 KΩ. D/SW 500/100 KΩ. D/SW 1 meg./100 KΩ. S/SW 500/500 KΩ. S/SW 500/500 KΩ. S/SW 500/1 meg.

Precision Tape Motors: £1-95 200/250V. Famous German manufacturer. Post 20p. Transistor Gang Condensers: 20p. Miniature AM. Post free. Modern Gang Condensers: 30p. AM/PM or AM only 20p. Post 10p. Transistors 15p each. Post free. AC126, AC128, AF114, AF117, OC45, OC71, OC81, OC81D. Valve ELL80 50p. Only stock in the country.

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#### PARKERS SHEET METAL FOLDING MACHINES **HEAVY VICE** MODELS

With Bevelled Former Bars	4	Carry, sall	Carrfree
No. 1. Capacity 18 gauge mild steel X : No. 2. Capacity 18 gauge mild steel X : No. 3. Capacity 16 gauge mild steel X : Some bench models. Capacities 36in.	24in. wide 18in. wide		£15 £10 £10 16 gauge £29.

End folding attachments for radio chassis. Tray and Box making for 36in. model, 27½p per ft. Other models 17½p. The two smaller models will form flanges. As supplied to Government Departments, Universities, Hospitals.

One year's guarantee. Money refunded if not satisfied. Send for details.

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#### FETS N CHANNEL FETS FULLY TESTED AND MARKED

Туре	BVgss (Min.)	igss (Max.)	Vp (Max.)	IDss (Max.)	R on (Max.)	Case
GP25*	20∨	InA	107	20mA	_	TO72
GP71	20V	InA	I2V	I50mA	190Ω	TO18

\* Similar to 2N3819

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ı, İ	CA3001	2.69 2.40		1 1.28 1.10
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1	CA3002	1.80 1.60	CA3037	1.65 1.47
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	CA3010	1.87 1.23	CA3043	1.87 1.21
	CA3010.		CA3044	1.20 1.07
ŀ	CA3011	0.74 0.65		1 1 20 1 0
1	CA3011		CA3045	1.28 1.09
		0.74 0.65	CA3046	0.69 0.60
	CA3012	0.89 0.79	CA3047	1.87 1.21
ı		V1 0-89 0-79	CA3047A	
H	CA3013	1.05 0.94	CA3048	2:04 1:81
	CA3014	1.24 1.10	CA3049	1.60 1.48
		V1 1 24 1 10	CA3050	1.84 1.84
	CA3015	2.09 1.86	CA3051	1.34 1.20
	CA3015.		CA3052	1.65 1.47
	CA3016	2.46 2.19	CA3053	0.46 0.41
		A 8.78 3.38	CA3054	1.09 0.97
	CA3018	0.84 0.75	CA3055	1.69 1.51
,		A 1.10 0.99	CA3056	1.20 1.07
	CA3019	0.84 0.75	CA3056A	
	CA3020	1.26 1.13	CA3059	1 65 1 46
	CA3020	A 1.60 1.48	CA3060	4.91 4.87
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	CA3022 CA3023	1.26 1.18		1.20 1.07
	CA3023	1.00 0.90	CA3066 CA3067	2·11 1·88 2·18 1·94
	CA3026		CA3068	2.48 2.16
	CA3020.	1.00 0.90	CA3000	1.70 1.51
	CA3028.		CA3071	1 62 1 44
	CA3028		CA3071	1.66 1.46
١	CA3029	0.87 0.77	CA3075	1 18 1 00
	CA3029		CA3076	1.80 1.16
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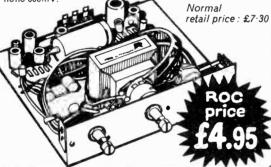
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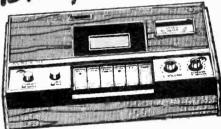
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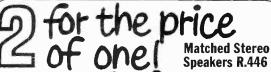
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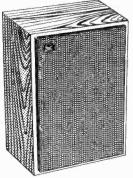


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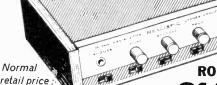
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10-watt Transistor Stereo Amplifier R.136

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8-2, 11, 13, 16, 24, 20, 100v. Micro Switches, \$/P, C/O. I Amp. Bridge Rect. 25v.	20 25 25	17 20 22	15 15 20

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B86	50	Sil. Diodes : IN914 and I		50p
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B60	10	7 Watt Zen Mixed Voltz		50p
H6	40	250mW. Ze DO-7 Min.		50p
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H15	30	Top Hat Sili 750mA. M	con Rectifiers, ixed volts	50p
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B79	4		07 Sil. I PIV lai				50p
B81	10		Switch and sn		ixed t	ypes	50p
B99	200		ed Capa rox. qu ht				50p
H4	250		rox. qu			ge 10p. ted by	50p
H7	40		wound and va				50p
Н8	4		7 Sil. P PIV. I		plastic	:	50p
H9	2		71 Ligh o Trans		sitive		50p
H12	50	NKT bran	155/259 d new s	Gerr tock	n, dio	des, nce	50p
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HI9	10		I/81D u type Pl			te	50p
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SI	EM	ıc	0	N	D	U	¢	т	0	R	s	
	- 17				12				-	۳.	,,	

AC126	12p	BFY52	22p	OC81	12p	2N3055	72p
AC127	12p	BSY56	30p	OC82	12p	2N3702	15p
AC12B	12p	BSX21	25p	ORP12	48p	2N3703	14p
ADI40	40p	BY124	71P	IN4001	7 <del>1</del> p 10p	2N3704	171p
AFI15	20p	BYZIO	20p	IN4002	ΙÓρ	2N3705	15p
AFII7	20p	BYZ13	20p	IN4003	Hp	2N3706	12p
BC107	10p	OA85	7p	I N4004	I2p	2N3707	181p
BC108	10p	OA91	5p	I N4005	13p	2N3708	IÕp
BC109	10p	OA202	7p	I N4006	13p	2N3709	Hp
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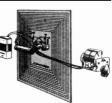
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Size approx. 17in  $\times$  10 $\frac{3}{4}$ in  $\times$  6 $\frac{3}{4}$ in. Drive unit 13in  $\times$  8in with parasitic tweeter. Max. power 10W, 3 ohms. Simulated Teak cabinet. £14 pair + £2 P. & P. Duo Type III Size approx.  $23\frac{1}{2}$ in  $\times$   $11\frac{1}{2}$ in  $\times$   $9\frac{1}{2}$ in. Drive unit  $13\frac{1}{2}$ in  $\times$   $8\frac{1}{4}$ in with H.F. speaker. Max. power 20W at 3 ohms. Frequency range 20Hz to 20kHz. Teak veneer cabinet. £32 pair + £3 P. & P.

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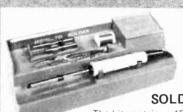
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BP47 = 7447	BCD-seven-segment decoder/drivers	2-00	1.75	1-50
	(15V outputs)	0.97	0-94	0-88
BP48 = 7448 BP50 = 7450	BCD-to-seven-segment decoder/driver	0.97	0.94	0.88
DF30 = 7450	Expandable dual 2-input and-or- invert	0-15	0.14	0.12
BP51 = 7451	Dual 2-wide 2-input and or-invert	0.10	0.74	0.12
DD10 - 7410	Quad 2-input expandable and-or-	0-15	0.14	0.12
BP53 = 7453	invert.	0.15	0.14	0-12
BP54 = 7454	4-wide 2 input and or invest gates	0.16	0.14	0.12
BP60 = 7460	Dual 4-input expander	0.15	0-14	0.12
BP70 = 7470 BP72 = 7472	Dual 4-input expander Single-phase J-K flip-flop Master slave J-K flip-flop Dual Master slave J-K flip-flop Dual D type flip-flop Quad latch Dual J-K with pre-net and clear	0.29	0-26 0-26	0-24
BP73 = 7473	Dual Master slave J-K flip-flop	0.37	0-25	0·24 0·32
BP74 = 7474	Dual D type flip-flop	0.37	0.35	0.82
BP75 = 7475 BP76 = 7476	Quad latch	0-47	0.45	0.42
BP80 = 7480	Quad latch Dual J-K with pre-set and clear Gated full adders 16-bit read/write memory 2-bit binary full adders Quad full adders	0.48	0·40 0·64	0-38 0-58
BP80 = 7480 BP81 = 7481	16-bit read/write memory	0.97	0.94	0-88
BP82 = 7482	2-bit bluary full adders	0.97	0.94	0.88
BP83 = 7483 BP86 = 7486	Quad full adder Quad 2-input exclusive NOR gates	1.10	1.05 0.30	0.95 0.28
BP86 = 7486 BP90 = 7490	BCD decade counter	0.67	0.84	0.58
BP91 = 7491	8-bit shift registers Divide-by-twelve counters	0-87	0.84	0.78
BP92 = 7492 BP93 = 7493	Divide-by-twelve counters	0-67 0-67	0.64	0-58 0-58
BP94 = 7494	Dual entry 4-bit shift register	0.77	0.74	0.68
BP95 = 7495	4-bit up-down shift register	0.77	0.74	0.68
BP96 = 7496	Quai 2 input exclusive NOR gates BCD decade counter 8-bit shift registers Divide by twelve counters 4-bit binary counters Dual entry 4-bit shift register 4-bit up-down shift register 5-bit parallel in parallel out shift- register	0.77	0-74	0-68
BP100 = 74100	8-bit bistable latches	1.75	1.65	1.55
BP104 = 74104	Single J.K flip-flop equivalent 9000			
BP105 = 74105	series Single J K flip flop equivalent 9001	0-97	9.94	0.88
	series	0.97	0.94	0.88
BP107 = 74107	Dual Master slave flip-flops	0.40	0.38	0.36
BP110 = 74110 BP111 = 74111	Dual data lock-out flip-flop	0.55	0.53 1.15	0.50 1.00
BP111 = 74111 BP118 = 74118	Hex set-reset latches	1.00	0.95	0.90
BP119 = 74119	Hex set-reset latches. 24-pin	1.35	1.25	1.10
BP121 = 74121 BP141 = 74141	series Dual Master slave flip-flops (tates master-slave flip-flops Dual data lock-out flip-flop Hex set-reset latches Hex set-reset	0.67	0·64 0·64	0-58 0-58
BP145 = 74145	BCD-to-decimal decoder/driver BCD-to-decimal decoder/driver O/C 16-bit data selector	1.60	1.40	1.30
BP150 = 74150	16-bit data selector	1.80	1.70	1.60
BP151 = 74151 BP153 = 74153	8-bit data selectors (with strobe)	1.00	0.95	0.90
BP154 = 74154	4- to 16-line decoder	1.80	1·10 1·70	0.95 1.60
BP155 = 74155	Dual 2 to 4 line decoder	1.40	1.30	1.20
BP156 = 74156 BP160 = 74160	Dual 2- to 4-line decoder O/C	1.40	1.30	1.20
BP161 = 74161	Sync. 4-bit binary counter	1.80	1·70 1·70	1.60 1.60
BP190 = 74190	16-bit data selector  8-bit data selectors (with stroke) Dual 4-line-to-1-line data  4- to 16-line decoder Dual 2- to 4-line decoder Dual 2- to 4-line decoder 0/c 8ync. decade counter 8ync. 4-bit binary counter 8ync. u-lown BCD counter 8ync. binary up-down counter (single clock line)	3.50	3.25	3.00
BP191 = 74191	Sync. binary up-down counter (single clock line)	3.50	8.25	8.00
BP192 = 74192	Sync. up-down decade counter	2.10	3·25 1·95	1.75
BP193 = 74193	clock line) Sync. up-down decade counter Sync. binary up-down counter (tow			
BP196 × 74104	clock lines)	2.10	1.95	1.75
BP196 = 74196 BP197 = 74197	Pre-setable 50MHz decade counter Pre-setable 50MHz binary counter	1-80 1-80	1·70 1·70	1.60 1.60
BP198 = 74198	Pre-setable 50MHz binary counter 8-bit parallel L-R shift register 8-bit parallel access shift register	5-50	5.00	4.00
BP199 = 74199	8-bit parallel access shift register	5-50	5.00	4-00
Devices may be n	nixed to qualify for quantity price. Larg	er quan	tities-p	rices on
Data is available f	'4 Series only). or the above series of I.C's in booklet for	n Prio	e 13n	
		1.10	Lop.	

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$UIC01 = 12 \times 7401N$	$50p$ UIC50 = $12 \times 7450N$	50p UIC82 = $5 \times 7482 \text{N } 50p$
$UIC02 = 12 \times 7402N$	$50p UIC51 = 12 \times 7451N$	$50p$ UIC83 = $5 \times 7483N$ $50p$
$UIC03 = 12 \times 7403N$	$50p$ UIC60 = $12 \times 7460N$	$50p \ UIC86 = 5 \times 7486N \ 50p$
$UIC04 = 12 \times 7404N$	$50p \ UIC70 = 8 \times 7470N$	$50p \ UIC90 = 5 \times 7490N \ 50p$
$UIC05 = 12 \times 7495N$	50p UIC72 = $8 \times 7472N$	50p UIC92 = $5 \times 7492N$ 50p
$U1C10 = 12 \times 7410N$	$50p$ UIC73 = $8 \times 7473N$	50p UIC93 = $5 \times 7492N$ 50p
$UIC20 = 12 \times 7420N$	50p UIC74 = 8 × 7474N	
		$50p UIC94 = 5 \times 7494N 50p$
$UIC40 = 12 \times 7440N$	$50p \ \ UIC75 = 8 \times 7475N$	$50p \ UIC95 = 5 \times 7495N \ 50p$
$UIC41 = 5 \times 7441AN$	$750p UIC76 = 8 \times 7476N$	$50p \ UIC96 = 5 \times 7496 N 50p$
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BP933	Dual 4-input expander		13p	120	11p
BP935	Expandable Hex Inverter		13p	12p	llp
BP936	Hex Inverter		13p	12p	11p
BP944	Dual 4-input NAND expandable buffer		t.		
	pull-up		13p	12p	11p
BP945	Master-slave JK or RS		25p	24p	22p
BP946	Quad, 2-input NAND		12p	11p	10p
BP948	Master-slave JK or RS		25p	24p	22p
BP951	Monostable		65p	60p	55p
BP962	Triple 3-input NAND		12p	11p	10p
BP9093	Dual Master-slave JK with separate clo		40o		
BP9094				38p	35p
	Dual Master-slave JK with separate clo		40p	38p	35p
BP9097	Dual Master-slave JK with Common Cl	ock	40p	38p	35p
BP9099	Dual Master-slave JK Common Clock		40p	38p	35p
	De mixed to qualify for quantity pri- (DTL 930 Series only).	ce. La	rger qu	antity p	rices on

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			Band)	53p	45p	40p
BP 709 -72709	D.I.L.	14	High OP Amp	53p	45p	40p
BP 709P µA709C	TO-5	8	High Gain OP Amp	53p	45p	40p
BP 711-4A711	TO-5	10	Dual comparator	58p	50p	45p
BP 741 -72741	D.I.L.	14	High Gain OP Amp	,		,
			(Protected)	75p	60p	50p
μΑ 703C-μΑ703C	TO-5	6	R.FI.F. Amp	43p	35p	27p
TAA 263—	TO-72	4	A.F. Amp	70p	60p	55p
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VOL. 7 No. 12 December 1971

## ELECTRONICS

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#### TODAY'S ICS AND TOMORROW'S

The profusion of integrated circuit devices on the retail market offers great opportunities for equipment designing and building. Yet the formidable lists of type numbers must be daunting to many a reader as he peruses those advertisements featuring i.c.s. Clearly there is a requirement for a guide that identifies the more common types, indicates function and provides basic application data. This need we have attempted to meet in the Linear and Logic IC Survey which is the subject of this month's special supplement.

It will be appreciated that this Survey is not, and indeed could not be, exhaustive. Only those types of integrated circuit known currently to be available via retailers are included. This in fact makes the Survey all the more valuable, since no reader is likely to be sent on a wild goose chase after some rare device that cannot be obtained, unless one has special connections

within the industry.

#### \* \* \*

Having (hopefully) clarified the current i.c. situation to some extent and produced two separate lists, one for linear, the other for logic devices, we learn that this orderly and simple segregation into just two clearly defined categories may not always be possible in the years to come. By strange chance, just as our Survey was completed news was released of what is claimed to be a revolutionary development in i.c. design and manufacture by a British firm. The Collector Diffusion Isolation (CDI) method now perfected by Ferranti permits both digital and linear circuits to be formed on the same monolithic chip; it combines the linear high performance and digital high speed capability of the bipolar method with the high circuit density capability offered by the MOS technique.

The developers of CDI have suggested that this new bipolar process will cause a general widening of MSI and LSI applications particularly in areas involving analogue-digital techniques; they have mentioned, specifically, desk calculators, fuel injection systems, washing machine controls, and model control as

examples where CDI will have great impact.

It is good to hear of a UK firm making an outstanding contribution to the development of new i.c. techniques. Just how important and significant the CDI method actually is, time will tell. But there are already two American contenders in the field. Both Fairchild and Raytheon have developed their own methods for increasing the component density on a single chip. Are we about to witness another price war? With mammoth production runs, technical considerations tend to play a minor role—the more economical process usually wins the day.

F.E.B.

#### THIS MONTH

#### CONSTRUCTIONAL PROJECTS

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TECH	INIQUES — 7	100

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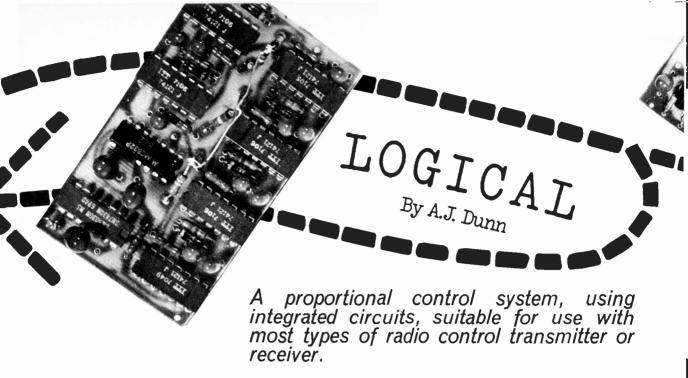
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#### SPECIAL SUPPLEMENT

LINEAR AND LOGIC IC SURVEY

Our January issue will be published on Friday, December 10.

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The vast number of integrated circuits now on the market has lowered the cost to levels whereby they can be used for many applications. Although the circuitry for the system described would be possible with discrete components, the complexity would be such as to deter many from contemplating a sophisticated system, whereas with integrated circuits, the modules are simple and easy to build.

This article deals with the coder and decoder sections only; a block diagram of a typical system is shown in Fig. 1a. These may be built and tested against each other (Fig. 1b) the interface or coupling units being simple stages allowing, for example, a transmitter with a -12V supply to be used with the coder which requires +5.2 to 5.7V.

The system is described for six channels although it can easily be varied for three to nine channels. It is not economic to have less than three channels; for more than nine channels some compromise is necessary with regard to proportionality and the time rate of control.

Printed circuit boards are necessary for this project and care should be taken to make these of a high standard since one interconnection short can cause complete malfunction and may be hard to trace; time taken wiring up is amply repaid.

Alternative forms of coder are described and it is not advisable to attempt construction without considering the corresponding decoder. Positive logic (output high = 1, low = 0) is used throughout

#### **PULSE CONTROL TECHNIQUES**

The control system involves the production of a train of (positive going) pulses, each pulse corresponding to a channel and being independently and proportionally variable in length from about 0.5 to 4ms.

As shown in Fig. 2a, the pulses occur sequentially but their position relevant to the start of the train is determined by the length of the other pulses and the interval time between pulses  $-\frac{1}{4}$ ms.

The pulse train is repeated at intervals  $\tau_1$  of 25 to 50ms or 20 to 40 times a second. This is the rate at which the pulse lengths are measured and changes detected. Allowing five measurements (but preferably ten) for the detection of a small change and servo initiation, this gives a system response time of about  $\frac{1}{8}$  to  $\frac{1}{4}$  second.

Due to the inertia of the motor and load, the servo has a response time which must be added to the above; a model aircraft flying at 60 m.p.h. would move 22ft in \(\frac{1}{4}\) second before, say, the rudder servo started operating.

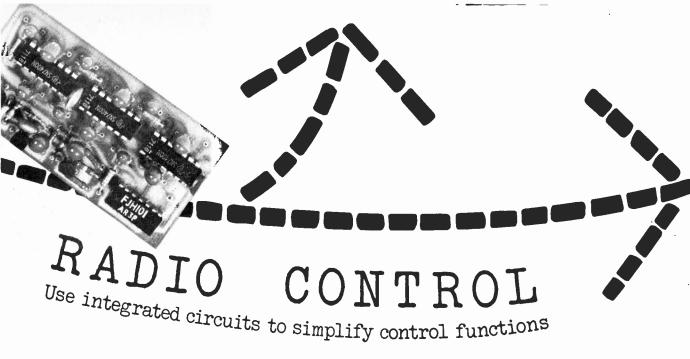
#### **PULSE LENGTH**

The pulse train repetition time  $\tau_1$  for n channels is made up of  $(n-1) \times \frac{1}{3}$ ms intervals, n pulses of maximum length  $\tau_3$ ms and a synchronising period  $\tau_2$  used in the decoder to determine the start of the pulse train. This synchronising period  $\tau_2$  must be  $1\frac{1}{2} \times \tau_3$  or preferably  $3 \times \tau_3$ .

A four-channel system with a maximum pulse length  $(\tau_3)$  of say 4ms would therefore give as a minimum  $(4 \times 4) + (3 \times \frac{1}{3}) + (1\frac{1}{2} \times 4) = 23$  milliseconds.

The sharp-edged pulses used in the coder and decoder are "softened" to approximately sine wave shape, before actual transmission and reception (as shown in Fig. 3), the pulse interval period being then approximately 0.5ms wide corresponding to a 2kHz tone. This allows a factor of approximately two for the bandwidth performance of any existing receiver, it being equally convenient to use a tone modulator transmitter or to use the i.c.w. mode and to detect the 27MHz carrier only in the receiver.

The maximum pulse length  $(\tau_3)$  is determined as a reasonable relationship with regard to the interval time, being restricted to contain  $\tau_1$  to less than 50ms. If a slower system response time can be tolerated, as in the case of a scale model boat, the maximum pulse length could be increased to 10ms or alternatively more channels could be used.



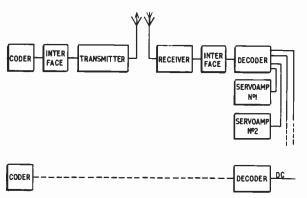


Fig. 1. Block diagram of a typical radio control installation

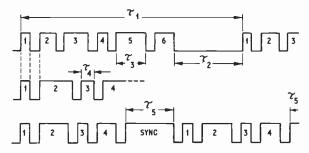


Fig. 2. Variable pulse lengths for proportional control

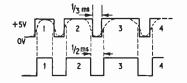


Fig. 3. The coder pulses are "softened" to almost sine wave shape for transmission to avoid troubles with short propagation times of square wave pulses

If only one channel, e.g. rudder, was required for maximum sensitivity, the corresponding maximum pulse length could be increased provided that the other channels in use were restricted operationally to prevent all the pulses being at maximum length, hence encroaching into the sync period  $\tau_{\rm p}$ .

#### CODER VARIANTS

Three variants of coder are described: coders 1 and 2A are similar in form providing alternative construction dependent upon the use of components that may be to hand. The cost of the timing capacitors becomes comparable with that of the "active" components. In these cases, the period  $\tau_1$  is fixed by using a multivibrator.

The coder 2B simply uses a longer positive-going pulse to provide the synchronising period shown as  $\tau_{3}$  in Fig. 2b; the channels are no longer being completely independent,  $\tau_{1}$  not being fixed. In this case the end of the synchronising pulse  $\tau_{5}$  is used to initiate the next pulse train.

The particular advantage is that, with some of the channel controls at minimum, a shorter system response time is possible. The disadvantage is that, operating any control, the corresponding pulse length is changed, in turn changing the total cycle time, and hence changing the measurement rate of the other pulses and their corresponding output.

Considering the case of a six-channel coder with all controls in the mid-position a pulse length of 2ms cycle time =  $(6 \times 2) + (6 \times \frac{1}{3}) + 6$  (synchro pulse) = 20ms. If one control is changed to maximum and its corresponding pulse length to 4ms, then the cycle time becomes 22ms.

The output signal derived from these pulses changes in the ratio  $\frac{2}{20}:\frac{4}{22}$  not 2:4 and the other channels outputs change by 20:22.

The greater the number of channels used, the less marked the loss of complete channel independence, and the greater the gain with respect to cycle and system response time.

Description	Logic type	Ferranti	Motorola	Mullard	sgs	Signetics	S.T.CI.T.T
Quad 2-input	TTL	ZN7400E	MC7400P	FJH131	6900259		
NAND gate	DTL	ZN346E	MC846		6994659		MIC9495D
8-input NAND gate	TTL	ZN7430E	MC7430P	FJHIOI		N7430A	
Dual JK flip-flop	TTL	ZN7473E	MC7473P	FJJ121		N7473A	
Monostable	TTL	ZN7470E		FJJ101		N7470A	
Retriggerable	TTL				9601		
Monostable	DTL				T118		
Dual 4-input NAND gate with node	DTL	ZN330E	MC830P		6993059		MIC9305D

#### **DECODER**

The decoder consists of two parts, the sync separator using a retriggerable monostable, and a shift register formed from a chain of JK flip-flops.

The received positive-going pulses repeatedly trigger the special monostable which has a full back-off time of at least  $1\frac{1}{2} \times$  the maximum signal pulse length. It can change state therefore only during the sync period, this change of state being used to produce a special pulse which is applied to all the "clear" inputs of the JK flip-flops.

During the sync period the outputs of the flipflops are set to "0", awaiting the train of signal pulses which are applied to all the "clock" inputs. One flip-flop is used to store a "1" which is successively moved down the chain (shift register) in steps coincident with the signal pulses.

The final individual output pulses from each flipflop output is equal to the period of the signal pulse together with the following interval (½ms). Either one more flip-flop than the number of channels is used (to store a "1") or the last output is gated out as shown, the gating being associated with the decoder or one servo unit as desired.

#### SUPPLY VOLTAGE

For reliable operation the supply should be constant and between + 5·2 and + 5·7V. The most convenient way of obtaining this voltage for the coder is by the use of a good 6V battery in series with a 150mA diode. Precaution should be taken to prevent more than 6V being applied even momentarily, a Zener diode being used with the decoder.

#### CODER I

The first coder is for six channels of positive going pulses. Four i.c.s only are used, one being an 8-input NAND gate (SN7430) to collate the outputs derived from an array of NAND gates (Fig. 4).

Pairs of these gates are used with a timing resistor and capacitor to give the signal pulse (for example, input 3 from G5 and G6 to G13) or the interval period (\frac{1}{3}ms) from G4 and G5.

The quad 2-input gate i.c. (SN7400) is used, therefore, to produce three timing periods and the series can be extended as desired. Eight channels would be a convenient limit unless G13 in IC7 is replaced by two or more gates followed by further grouping.

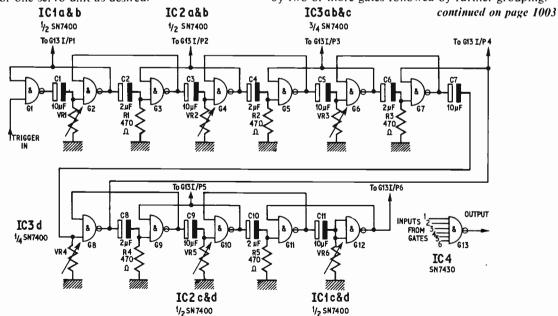


Fig. 4. The pulse lengths are determined and set by the potentiometers in the timing circuits. NAND gates are used to determine the timed sequence of pulses being fed to the master control gate G13

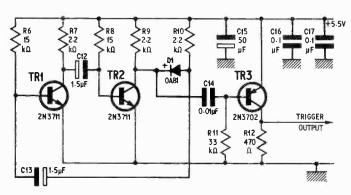
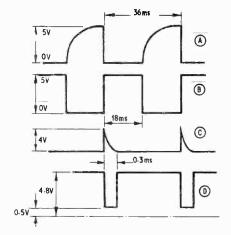


Fig. 5. Triggering is dependent on coincidence of an incoming clock pulse from this multivibrator and trigger circuit



#### COMPONENTS . . .

CODER I (Figs. 4, 5, and 8)

Resistors

RI to R5 470 $\Omega$  (5 off)

 $15k\Omega$ R6 R7

 $2.2k\Omega$ R8  $15k\Omega$ 

R9  $2.2k\Omega$ 

RIO  $2.2k\Omega$ 

RII 33k $\Omega$ 

R12 470Ω

All  $\pm$  5%,  $\frac{1}{4}$ W carbon

**Potentiometers** 

VRI to VR6  $2.5k\Omega$  for pre-selection of values, then replacement by fixed resistors (see text)

Capacitors

 $\dot{C}1$ , C3, C5, C7, C9, C11  $10\mu F$  tantalum or elect. 20V (6 off)

C2, C4, C6, C8, C10  $2\mu F$  polarised tantalum or elect. 25V (5 off)

C12, C13  $1.5\mu$ F elect. C14  $0.01\mu\text{F}$  disc ceramic C15  $50\mu\text{F}$  elect. 15V

C16, C17 0-1 µF disc ceramic (2 off)

Intergrated circuits

ICI, IC2, IC3 SN7400 quad 2-input NAND gate (3 off) IC4 SN7430 8-input

**Transistors** 

TRI, TR2 2N3711 npn silicon TR3 2N3702 pnp

NAND gate

silicon

Diode

DI OA81 or any 100mA signal diode

Miscellaneous

Fibreglass printed circuit board (see Fig. 7)

Fig. 6. The waveforms appearing at (a) TRI collector, (b) TR2 collector, (c) TR3 base, (d) TR3

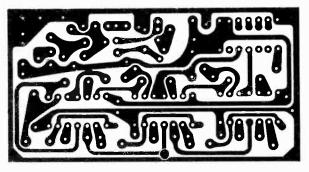
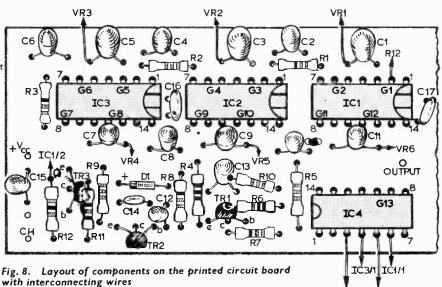


Fig. 7. The whole of "Coder I" and clock pulse generator is made up on fibreglass printed circuit board; the pattern here is reproduced full size



985

IC3/10 IC2/1

## PATENTS BEVIEW...

#### INFLAMMABLE LIQUID LEVEL MONITOR

AN interesting new use of fibre loptics occurs in BP 1 223 769 which is in the name of Maria and Giuseppe Panerai. Their Patent Specification is predominantly concerned with gauging the level of liquids in tanks.

Reading between the lines their main interest is in sensing the level of highly inflammable liquids such as petroleum where it is obviously far from ideal to use any electrical system in contact with the liquid. Of course, simple mechanical systems tend to be unreliable, especially if they are gravity dependent

What these Italian inventors suggest is to arrange a sequence of prisms or equivalents in a vertical line, up from the base of the liquid tank, so that as the tank fills more and more prisms are submerged.

The prisms protrude from a removable pipe, each one backing onto a photosensitive cell. A lamp is fixed to the top of the tank so as to illuminate the interior. It is quite easy to see that, as the liquid level goes down, so more and more prisms will be exposed and illuminated and more and more photocells activated. The cells can be linked to any simple electronic circuitry outside the tank by wires with a consequent minimal risk of sparking and explosion.

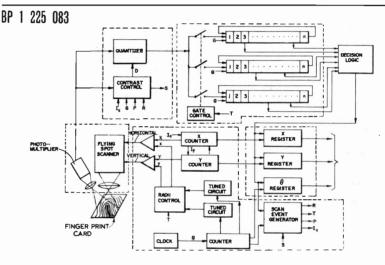
But none of this concerns fibre optics so far. It turns up almost as an afterthought in one of the

examples given in the specification where it is suggested that instead of putting the photocells directly behind the prisms, i.e. in the pipe which is submerged in the liquid, the photocells can be arranged elsewhere and the light from the prisms taken to them by light pipes or optic fibres.

#### AUTOMATIC FINGERPRINT IDENTITY

THE North American Rockwell Corporation have a new British Patent BP 1 225 083 which applies to the automation of fingerprint straightforward "off-the-shelf" electronics to provide means for scanning the fingerprint area. The major scan pattern has a minor scan pattern repeatedly superimposed over it.

The major scan pattern is a pre-determined linear pattern, which scans through a succession of relatively small area portions, and the minor pattern scans over each of these portions in a polar mode. This minor scan produces signals indicative of the pattern at that portion and these are stored in a bank of storage elements. The states of selected signals in these



checking. Rockwell start their specification with some sobering tacts.

For instance, the FBI in America has a fingerprint file of over 182 million fingerprint cards, each having ten prints on it. So, if checking a suspect's prints against the records is to be finished in his lifetime, sophisticated automation is obviously necessary. What's more, with lost or stolen credit cards being an escalating problem, it may well be necessary one day to identify a legitimate card-holder by his fingerprints at the time of purchase.

Rockwell remind us that recognising the minutiae of a fingerprint is basically a problem in pattern recognition, which is complicated by the obvious fact that the minutiae occur at arbitrary orientations.

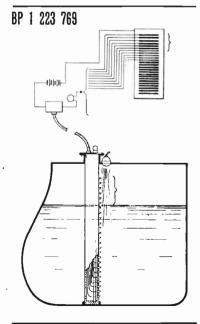
Whereas others have tended to concentrate on new techniques such as holography (matching prints with known masks), Rockwell propose a system which uses

storage elements are then sensed.

In more detail, Rockwell suggest the use of a flying spot scanner to derive an electrical analogue signal indicative of the fingerprint pattern at each small portion. This analogue signal is converted into digital form for storage and there is constant circulation of the signal in the memory through each of the storage elements so as to help the recognition of minutiae regardless of angular orientation.

An automatic contrast control circuit adjusts the detection process as a function of the general overall fingerprint image so that characteristics like ridge endings will produce distinctive light and dark areas to be encountered by the flying spot scanner moving in the orbits suggested.

Rockwell give plenty of details on suitable scanning techniques giving a total of 300,000 individual locations scanned on a single print. For present purposes, their block diagram gives the general picture.





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62	250			5 × 12					5.05	67
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92	1000			8 × 17					17-94	
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66	300	6	0 10	2 × 10	)·2 ~	9.5			3.38	52
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84	1000	16	0 11	4 × 14	4.0	14.0	11	4.1	9:12	82
93	1500	28	9 13	5 × 14	1.9 ×	16.5			13-22	
95	2000	40	0 17	8 × 16	5·5 ×	21.6		13	17-26	
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mai	mains lead and two 115V outlet sockets, £6-85, P & P 67np.									
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	Amps.			Size	cm,		Secondo	ary Wind	ings P	& P
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111 0.5	0.25		2 7·6 × 5	5.7 × 4.4	0-12V at	0-25A × 2	0.74	22		
213 1-0	0.5	1	0 8·3 × 5	5-1 × 5-1	0-12V at	0.5A × 2	0.88	27		
71 2	1			5.4 × 5.7	0-12V at	IA×2	1-16	2:		
18 4	2	2	4 8·3 × 7	7.0 × 7.0	0-12V at	2A × 2	1.62	36		
70 6	3	3 1	2 10·2 × 7	7.6 × 8.6	0-12V at	3A × 2	1.95	40		
72 10	5	6	3 7.9 × 10	0-8 × 10 2	0-12V at	5A × 2	2.56	5		
17 16	8	7	8 12·1 × 9	9.5 × 10.2	0-12V at	8A × 2	3.95	52		
115 20	10	11.1	3 12:1 × 11	1.4 × 10.2	0-12V at	10A × 2	5.03	6		
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226 60	30	34	0 17:0 × 14	4.5 × 12.5	0-12V at	30A × 2	17:05			
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No.		1b oz			€ Np
112	0.5	1 4	8-3 × 3-7 × 4-9	0-12-15-20-24-30V	0.88 22
79	1.0	2 0	7.0 × 6.4 × 6.0	**	1-18 36
3	2 0	3 2	8.9 × 7.0 × 7.6	11	1.75 36
20	3.0	4 6	10.2 × 8.9 × 8.6		2:16 42
21	4.0	6 0	10.2 × 10.0 × 8.6	11	2.56 52
51	5.0	6 8	12-1 × 10-0 × 8-6	0. 0	3-18 52
117	60	7 8	$12 \cdot 1 \times 10 \cdot 0 \times 10 \cdot 2$		3.79 52
89	10.0	12 2	14-0 × 10-2 > 11-4	11 11	6.21 67
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102	0.5	1 (1	7.0 × 7.0 × 5.7	0-19-25-33-40-50V	1.16	30
103	1.0	2 10	8-3 > 7-3 × 7-0		1-69	36
104	2.0	5 0	10.2 × 8.9 × 8.6		2.34	47
105	3.0	6 0	10.2 × 10.2 × 8:3	11	3 18	52
106	4.0	9 4	12-1 × 11-4 × 10-2		4.20	52
107	6.0	12 4	$12 \cdot 1 \times 11 \cdot 1 \times 13 \cdot 3$	11	6.21	67
118	8.0	18 9	13-3 × 13-3 × 12-1	31 11	8.10	97
119	10.0	19 12	165 × 11-4 × 15-9		10:15	97
Ref.	Amps.	Weight	Size cm.	60 VOLT RANG	E P	& F
No.	•	1b oz			£	NE
124	0.5	2 4	8·3 × 9·5 × 6·7	0-24-30-40-48-60V	1.18	36
126	1.0	3 0	8-9 × 7-6 × 7-6		1.64	36
127	2.0	5 6	10-2 × 8-9 × 8-6	15 11	2.56	47
125	3.0	8 8	11.9 × 9.5 × 10.0		3.90	57

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5	4.0		10-2 × 7-0 - 8-3		1.77	42
86	6.0	5 12	10·2 × 8·9 × 8·3 }	units do not in-	2.67	52
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#### By K. J. Matthews M.Sc.

LUMINESCENCE is the general term which is applied to the production of light, either visible or infrared, by the direct conversion of some other form of energy. The general term is then subdivided to denote the particular energy conversion involved.

Therefore, the production of light from heat, an effect which has been known for some considerable time, is designated thermoluminescence. A more modern, but very popular effect, is cathodoluminescence. This is light from kinetic energy or particle bombardment and keeps millions of T.V. screens active.

#### **ELECTROLUMINESCENCE**

Other forms include chemiluminescence, and bioluminescence, which is a division of chemiluminescence and occurs in some deep sea fish and in glowworms. Photoluminescence is a special case as it involves a light to light conversion, the emitted light differing from the stimulating radiation in frequency.

Finally, electroluminescence is the emission of light by direct conversion of electrical energy. The standard abbreviation for the effect is written EL. This process is not to be confused with the ordinary tungsten filament bulb, where there is an intermediate stage of heat making the process thermoluminescent.

A few EL devices are now available to the public. Unfortunately these are as yet only infra-red emitters, but visible light devices do exist and are certain to become available in the near future.

#### **USES**

Uses for these devices are self evident. In most cases they can be substituted wherever small panel lamps are employed, and their higher efficiency will

make them suitable for use with battery powered equipment. Numerical displays have been fabricated and because conventional read-out tubes require high voltages, EL devices of this type are likely to become very popular.

At the present stage in the development of El. devices, their use for general lighting seems a little remote. They cannot compete with fluorescent lighting for efficiency and only give monochromatic light.

It is not the aim of this article to describe practical applications but to give some insight into the mechanism of EL and of trends in present day research.

#### SEMICONDUCTOR BAND STRUCTURE

Some knowledge of semiconductor band theory is necessary before any understanding of electroluminescence is possible. Unfortunately this can be very complex and it is only possible to scratch the surface of the topic.

Fig. 1 depicts a semiconductor in energy band form. It consists of two bands, the conduction band and the valence band with an energy gap separating them. Electrons cannot exist with energies within this gap, and for this reason it is known as the forbidden gap or region.

For an electron to move from a valence band to a conduction band, it must "jump" the energy gap. A gradual transition is not allowed.

If the semiconductor has no energy, the conduction band will be void of electrons and the valence band full. In this state it is electrically inert—an insulator. However, a semiconductor always has some energy by virtue of its temperature, and this causes electrons to be propelled across the gap into the conduction band.

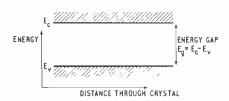


Fig. 1. Simple energy band diagram of a semiconductor

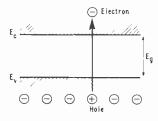


Fig. 2. The excitation process

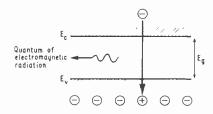


Fig. 3. Radiative recombination

To do this it is necessary that an electron receives an amount of energy at least equal to the energy represented by the width of the forbidden gap to enable it to make the jump. If it receives less it will not be able to do so. As we will see later, the size of the gap is constant for a particular semiconductor and is measured in terms of energy, in electron-volts (eV).

#### **EXCITATION**

When an electron receives sufficient energy to jump the gap to the conduction band, it leaves behind in the valence band a vacancy known as a "hole." See Fig. 2. It is convenient to think of these holes as positive electrons, but with reduced mobility.

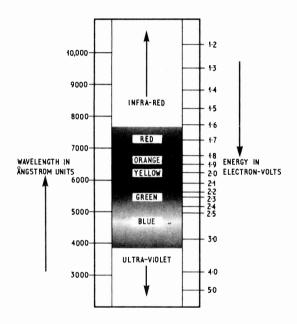


Fig. 4. Wavelength to energy conversion chart

If energy is now injected into the semiconductor for a short time and then removed, the internal equilibrium of the semiconductor is upset. There will be too many electrons in the conduction band and the semiconductor is said to be in an excited state.

There are many ways in which equilibrium may be restored. Here are three of the more important processes:

(1) the electrons may be ejected from the material entirely—similar to thermionic emission.

(2) the electrons may impart their energy to the crystal lattice of the semiconductor causing its atoms to vibrate. These vibrations are known as phonons.

(3) the electrons may fall back across the gap to occupy holes in the valence band, emitting as it does so, quanta of electromagnetic radiation, the frequency of which is determined by the width of the forbidden gap. This process, known as electron-hole recombination, is illustrated in Fig. 3.

These three processes, and others unmentioned, all occur in any semiconducting material simultaneously competing with each other. It is obviously the latter radiative process which is of interest to

EL, and it is desired that this be the main process if good light emitting efficiencies are to be obtained.

#### CRITERIA FOR ELECTROLUMINESCENT MATERIAL

In some semiconducting materials a radiative transition can only be achieved after a process of the phonon type has occurred. This is a characteristic of the material and cannot be altered. As this process is a two stage one, the energy gap of the semiconductor is known as indirect, or the material is called an indirect semiconductor.

In other materials, known consequently as direct semiconductors, the radiative transition occurs singly. Other processes still compete but these are minor compared to the serious drawback of an indirect energy gap. They can in any case often be minimised by reducing the temperature of operation of the semiconductor if super-efficiency is required.

Clearly then, the first criterion for an electroluminescent material is that it be direct. The second is the width of the energy gap, as this determines the frequency of the emitted light.

According to quantum physics, energy and frequency of light are related by the equation.

Energy =  $h \times frequency$ 

where h is Planks constant =  $6.62 \times 10^{-27}$ 

Since frequency =  $\frac{\text{velocity}}{\text{wavelength}}$ 

the energy equation becomes

Energy =  $\frac{h \times \text{velocity of light}}{\text{wavelength of light}}$ 

Substituting for h and velocity of light and converting to convenient units yields

Wavelength in angstrom units (Å) = 12,400Energy in electron volts (eV)

The visible spectrum runs from 8,000Å (red) to 4,000Å (blue) which on conversion gives red as equivalent to 1.55eV and blue to 3.0eV.

To recapitulate, the semiconductor must have a direct gap, and this must lie within the range, 1.55 to 3.00eV for visible light emission. If the gap is narrower than 1.55eV, infra-red emission will be obtained. See Fig. 4.

#### **DEVICES**

Before investigating the available EL materials consideration must be given to the method of its use in the production of a useful device.

As we have seen the trick lies in the method by which non-equilibrium conditions are set up in the semiconducting material. This can be done by any form of energy input, but for EL it must obviously be brought about by electrical energy. There are two known ways of doing this, and this causes a split in the research into EL devices as the resulting products are so unalike.

The older method is the electroluminescent panel in which excitation is brought about by a high electric field although the full theory is not understood.

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32/450V	20p	8+8/450V	18p	32+3	2/450	V	33 p
25/25V	10p	8+16/450V	20p	350 + 5			50p
50/50V	10p	16+16/450V	25p	32+32			
100/25V	10p	32 + 32/350V	25p	100 + 5			
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PAPER 350	V-0 1	4p, 0.5 13p;	mF 1	5n · 2n	F 150	V 1F	n ap.
500V-0:001	to 0.0	5 4p; 0 1 5p;	0.25	Sn O	47 95		ψ.
011 TITE 11		v , o x op,	2 20	ωp, υ		,,	

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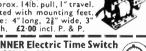
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45 185 230 280 600	9-12 9-12 18-32		73p* 78p* 73p* 78p*	700 700 1,250 2,500 2,400	24-36 36-45 30-48	2 c/o HD 6M 4 c/o 6M 4 c/o	73p* 63p* 63p* 63p* 50p	
700	16-24	4M 2B			40-70	4 c/o	63p*	
700	16-24	4 c/o	78p*	9,000	40-70	2 c/o	50p*	
700	12-24	2 c/o	63p*	15k	85-110	6M	50p*	
(a) c	oil ohm	s: (2) Wo	rking	t.c. vol	rs: (3) Co	ntacts: (4)	Price	

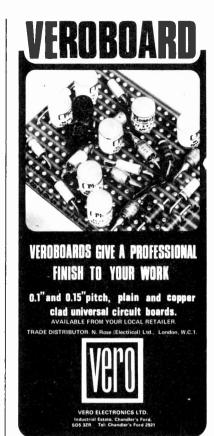
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This method has been known for some time but modern materials research has improved the EL panel vastly, making d.c. operation feasible where before only a.c. was possible.

#### STRUCTURE OF PANEL

The basic construction of a panel is simple, see Fig. 5, although more complicated, and more efficient geometries have been produced. In a d.c. display, the backing electrode which is the cathode is made of aluminium.

The anode, through which the light must pass, is of glass treated in such a way as to make it conductive. This is accomplished by placing a very thin coating of tin oxide on its surface, by vacuum evaporation or by a chemical method, so that it remains transparent.

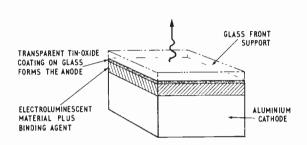


Fig. 5. Structure for a simple type of EL panel

The electroluminscent material is then mixed with a binding agent and spread very thinly between these electrodes, so that a high field is produced by a comparatively low voltage, typically 50 microns for 20 to 30 volt operation.

A newer method of stimulating EL and one which has been causing a great deal of interest, is based on the pn junction. The theory of this method of operation is better understood than the EL panel which still presents large areas of mystery.

If we suppose that a suitable material is available in both forms, i.e. p- and n-type, then a pn junction may be formed to make a diode.

In any junction diode, recombination of electrons and holes occurs at the junction when the diode is conducting in the forward direction as this is basic to its operation. If the semiconductor favours radiative recombination then the result will be light emission from the junction, produced directly by the passage of electric current through the diode.

To enable the light to escape the device, special geometries have to be employed in the manufacture of a crystal lamp if the bulk of the material is not to re-absorb it. Basically this means making a large area junction and arranging the emitting surface so that total internal reflection does not occur, which is by no means an easy job with the high refractive indices of most semiconductors. One type of arrangement is shown in Fig. 6.

#### MATERIALS FOR ELECTROLUMINESCENCE

How many materials are useful or potentially useful semiconductors? Germanium and silicon certainly and gallium arsenide is by now fairly well

known, but the list doesn't stop there. Hundreds more are possible of which only a few as yet have been investigated. It would seem therefore that suitable EL materials would be commonplace but, unfortunately, this does not appear to be the case.

In the crystal lamp field if not in EL panels there is a desperate need for a good semiconductor. This shortage has led to the employment of one or two sophisticated tricks but the problem is by no means satisfactorily solved.

Consider the common semiconductors to see how they conform to the criterion laid down earlier. Germanium—indirect gap of 0.66eV, silicon—indirect gap of 1.08eV, both clearly of little use being so far in the infra-red as to make them inconvenient even if they were direct.

Gallium arsenide however is clearly much more

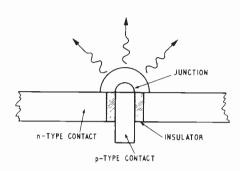


Fig. 6. Hemispherical structure of crystal lamp

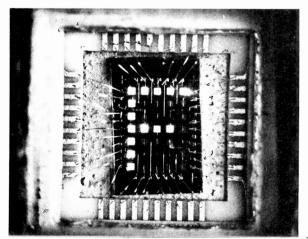
promising, having a direct gap of 1.5eV, tantalisingly close to visible. The available devices, 56CAY, MGA100, are made from gallium arsenide.

#### PERIODIC TABLE

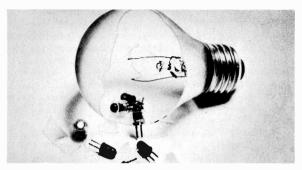
Why is gallium arsenide a semiconductor? A glance at a section of the periodic table of the elements, Fig. 7, throws some light on this question and indicates where we may look for additional material.

We see that germanium and silicon lie in group IV as do tin and carbon. As diamond can be regarded as a semiconductor and so, under certain

An array of 35 gallium phosphide lamps within a package overall size  $\frac{1}{4}$  in  $\times$   $\frac{1}{4}$  in approximately. In this photograph, the lamps have been illuminated to form the initial letter of the maker's name, Ferranti



Practical Electronics December 1971



Plastic encapsulated gallium arsenide phosphide red light emitting diodes made by Hewlett-Packard. These solid state lamps are designed for panel mounting or for printed circuit board mounting

conditions, can tin, there seems to be something about group IV elements which have some connection with semiconducting properties.

If one element from group III is combined with one from group V, this also produces a semiconductor. These are known as compound semiconductors or compounds and gallium arsenide falls into this class.

п	ш	IV.	¥	VI.
	В	С	N	0
	Al	Si	P	S
Zn	Ga	Ge	As	Se
Çd	In	Sn	Sb	Te
Hg	Τι	РЬ	Bi	Po

Fig. 7. Section of the Periodic table of the elements

In a similar fashion II-VI compounds are possible. Thus many semiconducting materials are possible and the more well known ones are listed in Table I. with their relevant parameters.

Confining ourselves to direct gap materials restricts the field considerably. All of the group IV elements are excluded, as well as three of the III-V compounds, the potentially useful three moreover. This leaves gallium arsenide as the compound with the largest direct gap, but this as we have seen is not good enough for visible EL.

At first sight the II-VI compounds are much more promising, zinc and cadmium sulphides both seem ideal, and indeed they are employed in the EL panel form of device. These materials, however, resist all attempts to grow them in both p and n-type forms and therefore crystal lamps cannot be constructed from them.

#### DOPING

The position for visible light crystal lamps would appear then to be hopeless, but two tricks are possible which partially solve the problem. Firstly, an indirect gap material with a larger than necessary gap can be used. Doping agents can be added to them which will create energy levels in the forbidden gap, effectively reducing it in size and giving it the characteristics of a direct gap.

Gallium phosphide doping in this way with zinc and oxygen will give a red emitting diode and the highest efficiency that has recently been obtained is in the order of 7 per cent. Inefficient green luminescence may also be produced from gallium phosphide, of approximately 0.1 per cent but this is

stretching the material's capabilities. Gallium phosphide of the high quality needed cannot, as yet, be consistently produced.

#### **MIXED CRYSTALS**

The second trick consists of the formation of a mixed crystal. Most of the III-V compounds are intermixable, that is, they can be grown together as a single, homogenous crystal. If in the growing of gallium arsenide for example, a proportion of phosphorus is added to the molten material, the resultant crystal contains a percentage of gallium phosphide.

This is a mixed crystal and should not be considered to be a mixture of gallium arsenide and gallium phosphide, but as a crystal of gallium arsenide phosphide, written Ga  $As_x P_{1-x}$ , the x indicating a whole range of compositions from pure gallium arsenide to pure gallium phosphide.

As gallium arsenide is direct and gallium phosphide is indirect, somewhere in the mixed crystal composition there must be a change over from the one type of de-excitation to the other, and it is possible to increase the energy gap for a considerable

way before it becomes indirect.

Red emitting lamps from gallium arsenide phosphide are possible if the crystal is grown with the correct composition. These lamps are becoming quite common in professional areas and it must only be a mater of time before they are generally available.

This is the state of the art at present as regards the crystal lamp field. Several other mixed systems have been tried and are workable propositions but they still only produce red light diodes. Present research is searching, rather desperately, for an efficient green emitter.

Table 1: PROPERTIES OF THE BETTER KNOWN SEMICONDUCTORS

Material	Туре	Formula	E <sub>g</sub> (eV)	De-Excitation Process
Diamond	IV	С	5.2	indirect
Silicon	IV	Si	1.08	indirect
Germanium	IV	Ge	0.66	indirect
Silicon Carbide	IV	SiC	2.9	indirect
Gallium Arsenide Gallium	III-V	GaAs	1.5	direct
Phosphide Gallium	III-V	GaP	2.25	indirect
Antimonide	III–V	GaSb	0.67	direct
Indium Arsenide	III-V	InAs	0.36	direct
Phosphide Indium	III–V	InP	1.29	direct
Antimonide	III-V	InSb	0.17	direct
Aluminium				
Phosphide Aluminium	III–V	AIP	3.00	indirect
Antimonide	III-V	AISb	1.62	indirect
Zinc sulphide Cadmium	II-VI	ZnS	3.7	direct
Sulphide Cadmium	II-VI	CdS	2.6	direct
Telluride	II-VI	CdTe	1.45	direct
Cadmium		,5016		direct
Selenide	II-VI	CdSe	1.74	direct

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2N697	18p	2N2925	22p	AC126	20p	BC154	20p	BFY51	20p
2N706	12p	2N2926	HP	ACI27	20p	BC157	I2p	BFY52	23p
2N930	29p	2N3053	27p	ACI28	20p	BC158	ilip	BSX20	I6p
2N1131	29p	2N3055	60p	ACI53K	22p	BC159	12p	C407	17p
2N1132	29p	2N3702	13p	ACI76	16p	BC167	ما ا	MC140	25p
2N1302	19p	2N3703	13p	ACY20	20p	BC168	10p	MPS6531	35p
2N I 303	19p	2N3704	13p	ACY22	16p	BC169	Hp	MPS6534	30p
2N1304	26p	2N3705	13p	ADI40	63p	BC177	14p	NKT211	25p
2N1305	26p	2N3706	13p	ADI42	50p	BC178	13 p	NKT212	25p
2N I 306	33p	2N3707	13p	ADI49	58p	BC179	14p	NKT214	23p
2N1307	33p	2N3708	10p	ADI6I	33 p	BC182L	Πp	NKT274	18p
2N1308	36 p	2N3709	Hp	AD162	36p	BC183L	10p	NKT403	65p
2N1309	36p	2N3710	13p	AFII4	24p	BC184L	Hp	NKT405	79p
2N1613	23p	2N3711	13p	AFI15	24p	BC212L	16p	OC71	38p
2N1711	26p	2N3819	23p	AFI17	22p	BC213L	16p	OC8I	25p
2N1893	54p	2N3904	35p	AFI24	24p	BC214L	16p	OCSID	25p
2N2147	95p	2N3906	35p	AFI27	22p	BCY70	18p	ZTX300	14p
2N2218	34p	2N4058	I3p	AFI39	33p	BCY71	33p	ZTX301	16p
2N2218A	44p	2N4059	10p	AF239	36p	BCY72	15p	ZTX302	22p
2N2219	38p	2N4060	Hp	ASY26	27p	BF115	23p	ZTX303	22p
2N2219A	53p	2N4061	Πp	ASY28	27p	BF167	18p	ZTX304	27p
2N2270	62p	2N4062	12p	BC107	12p	BFI73	19p	ZTX500	18p
2N2369A	19p	2N4124	18p	BC108	Пр	BF194	14p	ZTX501	21p
2N2483	35p	2N4126	27p	BC109	12p	BF195	15p		
2N2484	42p	2N4284	15p	BC125	15p	BFX29	31p	ZTX502	25p
2N2646	47p	2N4286	15p	BC126	22p	BFX84	25p	ZTX503	22p
2N2904A	42p	2N4289	15p	BCI47	10p	BFX85	52p	ZTX504	52p

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Code	Power	Tolerance	Range	Values available	I to 9 (see no	10 to 99 ste below)	100 up
0 3 3 3 3 3 3 3 3 3 3 3 3 3 3 3 3 3 3 3	1/20W 1/8W 1/4W 1/2W 1W 1/2W 1/2W 1/3W 7W	5% 10% 5% 10% 2% 10%±1/20Ω 5%	82Ω-220ΚΩ 4-7Ω-470ΚΩ 4-7Ω-10ΜΩ 4-7Ω-10ΜΩ 4-7Ω-10ΜΩ 10Ω-1ΜΩ 0-22Ω-3-9Ω 12Ω-10ΚΩ	E12 E24 E12 E24 E12 E24 E12 E12	9 1 1 1·2 2·5 4 7 7	8 0.8 1 2 3.5 7	7 0-7 0-7 0-9 1-9 3 6 6

Codes: C = carbon film, high stability, low noise.

MO = metal oxide, Electrosil TR5, ultra low noise,
WW = wire wound, Plessey.

Values: E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

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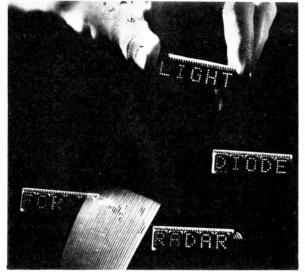
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The crystal systems  $\ln_x Ga_{1-x}P$  and  $\ln_x A11_{1-x}P$  are the most promising materials as yet investigated as they can provide direct energy gaps of the order of  $2\cdot 2eV$ . These materials are exceptionally difficult to grow, however, and while  $\ln P$ -GaP can be grown to the right composition,  $\ln P$ -A1P has not yet been produced.

Even if these materials were available, they only give green, what do we do to obtain blue? At this stage we enter the realms of conjecture, the nitrogen III-V compounds, largly unknown, appear to hold out some promise although they have a different crystal structure to all of the other semiconductors mentioned here. Aluminium nitride and gallium nitride appear to be the favourites for investigation.

#### FOR THE FUTURE

With their limitations therefore, both crystal lamps and EL panels are working propositions. The red crystal lamp is quite well established and will undoubtedly become a part of the vast range of common semiconducting devices, but the EL panel is held up by one snag—its short working lifetime. This is caused by impurities slowly diffusing into the luminescent material and is a major problem, which until it can be overcome, will limit its usefulness seriously.

The "flat" television screen seems to be a very long term project indeed. A panel's electrodes can be patterned so that any particular part of the screen may be addressed in a similar manner to a computer memory matrix, but the definition possible by this method is severely limited and in any case only one colour is possible.

When green and blue crystal lamps are available perhaps this will provide the answer, but the interconnection problems with such a vast number of individual elements is enough to make even the bravest engineer apprehensive.

However, these problems are not insurmountable since the Sony Corporation have shown, by presenting full colour TV on a large screen consisting of a vast array of red, green and blue light bulbs. It remains therefore to obtain the devices and then await the consequences.



If you do your own photo printing; in colour or monochrome, this three-in-one unit will go a long way towards ensuring reliable results. It combines the assessment of tonal balance with stabilised lamp voltage and accurate timing to produce first-class prints from your best negatives

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N this the second and final part, constructional details for the electronic ignition system are given together with installation instructions and testing procedures, for both positive and negative earth cars.

#### CONSTRUCTION

The prototype system was housed in a  $7\frac{1}{4}$ in  $\times$   $4\frac{1}{2}$ in  $\times$  2in Eddystone die-cast box. All the components are mounted on the printed circuit board (Fig. 11) with the exception of the transformer, TR4, TR5 and the thyristor. These are bolted to the case which acts as a heatsink.

#### PRINTED CIRCUIT BOARD

It is strongly recommended that turret tags be used to make connection to the printed circuit board and these should be inserted and soldered before any of the components are mounted. Fig. 12 shows the component locations.

#### INVERTER TRANSFORMER

The inverter transformer T1 is wound on a Mullard FX2243 pot core with a DT2206 bobbin and should preferably be impregnated when finished to keep out moisture which could cause leakage or damage to the windings of the transformer.

The first job is to wind on 400 turns of 34 s.w.g. enamelled wire in eight neat layers of fifty turns, with thin insulation between each layer. Take great care not to cross adjacent turns. This high voltage winding is now insulated from the others with about four layers of insulation tape.

Now wind on 12 turns of 20 s.w.g. bifilar (i.e. two wires wound together) in two layers, and lastly the three turns of 30 s.w.g. bifilar feedback winding. The wires should be colour coded with sleeves as shown in Fig. 13 and another layer of insulation tape wrapped around the bobbin. Try to ensure that the sleeves go as far as possible into the bobbin so that when it is placed inside the ferrite cups the enamel insulation is not scratched by the core.

If the core cannot be impregnated then it is recommended that the faces are firmly joined with a thin layer of Araldite or the transformer may shriek objectionably. This does not happen of course if the transformer is impregnated.

The reliability of the whole system depends on the care with which the transformer is made.

#### **UNIT ASSEMBLY**

The box should be drilled as shown in Fig. 14 and the holes deburred with larger drills, taking care to remove any roughness where the mica washers are to be placed for TR4, TR5 and the thyristor.

All the components can now be assembled in the box ready for the final wiring. Cadmium plated nuts and bolts should be used, with the exception of the bolt securing the thyristor, since steel ones will soon rust.

Interwiring details for the ignition system are

given in Fig. 15.

The thyristor should be mounted with a nylon bolt to remove the possibility of any flash over at this point due to the high voltages. The transistors



By D. S. GIBBS and I. M. SHAW (Ferranti Ltd)



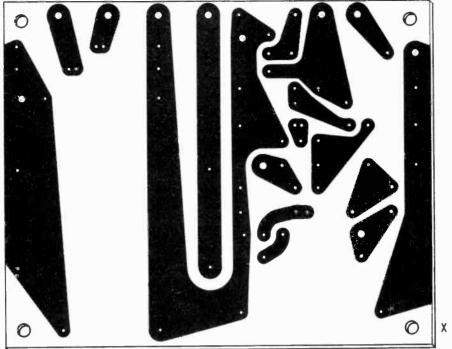


Fig. 11. Printed circuit board for ignition system, shown full size

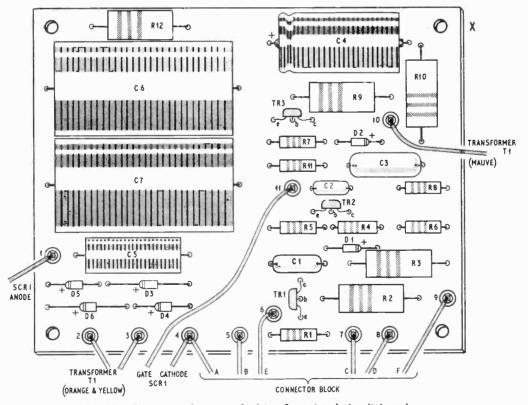


Fig. 12. Component layout and wiring for printed circuit board

### ... reliability depends on careful transformer construction



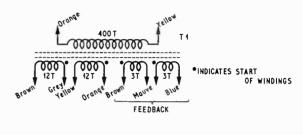


Fig. 13. Winding details, for inverter transformer

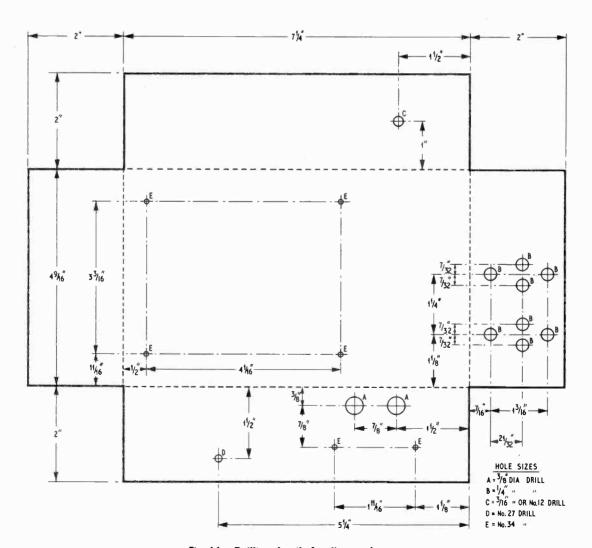


Fig. 14. Drilling details for die-cast box

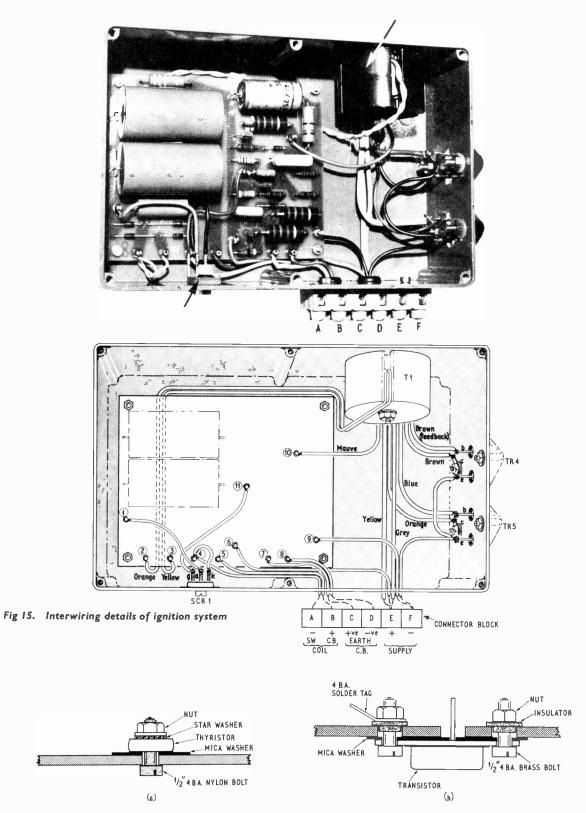


Fig. 16. (a) Mounting details for thyristor. (b) mounting details for inverter transistors



Plastic covers are fitted over the inverter transistors to prevent accidental short circuits

TR4 and TR5, and the thyristor, must be mounted on mica insulating washers as indicated in Fig. 16 with solder tags on the lower two transistor fixing screws for the collector connections. Plastic covers should be fixed to the transistors to prevent accidental short circuits when the unit is installed in the car.

The printed circuit board is mounted on spacers of about 4 in in length. Use a spring washer between the transformer mounting bolt and the core so that the transformer can be secured firmly without danger of cracking the ferrite core.

The unit can now be wired with 14/0076 wire in assorted colours for identification. Care must be taken to connect the transformer windings correctly or the inverter will not oscillate.

The wires to the six-way connector block are brought out via two grommets just below it.

#### CHECKING THE UNIT

After checking the wiring carefully some simple checks should be performed before installation, if the constructor does not have facilities to fully test the operation of the unit.

If the construction has been carried out as described using only good quality components the only check necessary is that the inverter transformer windings have been correctly connected and that it delivers the correct output voltage.

To do this connect the car battery between (E) and (F) on the connector block. The transformer should be heard to "sing" immediately and approximately 500 volts d.c. can be measured between tags I and 4 on the printed circuit board (positive on tag I). If this is not correct then the phases of the transformer windings will have to be examined and corrected before installation.

The current taken from the battery is approximately 500mA when the inverter is operating correctly.

#### **INSTALLATION**

All connections from the car for both positive and negative earth are made to the connector block, but since there are some differences they will be explained separately.

First, a suitable place in the engine compartment should be found to mount the unit so that the wires can be cut to the correct lengths. In both systems the wire connecting the ignition coil to the contact breaker must be removed but kept, so that the car can be reconstructed to the conventional system if necessary at any time.

#### POSITIVE EARTH CARS

The wire or wires connected to the SW or negative terminal of the ignition coil for positive earth systems, must be connected instead to the negative supply input to the unit (F) since this supply is via the ignition switch.

The two connections between the unit and the coil are made next; SW or negative to (A) and CB or positive to (B). These must be the only connections to the coil terminals.

The contact breaker is connected to (C) and good earth from the car chassis or positive battery terminal to (E). There are a total of five wires to the unit for positive earth cars with contact (D) left disconnected.

#### **NEGATIVE EARTH CARS**

The connections for negative earth cars are somewhat simpler since the wires to the positive terminal of the coil can be left connected and the positive supply terminal of the unit (E) connected to it.

The contact breaker is connected to (D) and the negative coil terminal to (B). Contact (F) is earthed making a total of four wires to the unit with contacts (A) and (C) left unconnected.

The astute reader will deduce that the first half cycle of the spark in the negative earth system is positive and not negative as is usually recommended. This can be corrected by reversing the connections to the coil. In some cases this may be inconvenient as the positive terminal is sometimes used as a junction point for leads from other equipment and the coil is supplied with more than one tag on the positive side to accommodate them. It is not really necessary to go to the trouble of reversing the initial spark polarity since the unit generates a dual polarity spark (Fig. 6a, Part One).

The unit can now be fitted to a convenient place in the engine compartment by drilling at least two suitable holes in the bottom of the box, and the car, and screwing the unit in place with self tapping screws.

#### COLD START COILS

Some cars are fitted with a "cold start" coil, which is a low voltage coil with a series ballast resistor that is shorted out when starting, to give an increased spark voltage under cold starting conditions. The ballast resistor is usually attached to one of the terminals of the coil but it may take the form of a resistance cable between the ignition switch and the coil.

. . . . never attempt to remove spark plug leads while engine is running

It does not actually make any significant difference to the spark produced by the unit whether the resistor is left in series with the coil or not, but the resistance cable should not be used to supply power to the unit as there will be a considerable loss of voltage along it. In this case it is best to remove the resistance cable and replace it with a length of wire. In all cases the connection between the coil and the starter solenoid must be removed.

#### **PRECAUTIONS**

Because of the high open circuit voltage produced by this unit *never* attempt to remove the spark plug leads whilst the engine is running, since not only is there a danger of breaking down the insulation of the coil but the experimenter stands a good chance of getting a nasty shock.

In addition the thyristor can be exposed to transient voltages of up to 1,000V under these conditions

and may be destroyed.

#### **PERFORMANCE**

In most capacitor discharge ignition systems the usual fault is for the thyristor to "latch on" with the short circuit current available from the inverter, especially when an iron cored transformer is used. This cannot occur in the system described here as explained in Part One.

The spark gap of the plugs may be increased with advantage for smoother idling, but not above 0.04in or the insulation of the ignition coil will be unduly taxed, especially in cars with high compres-

sion engines (> 10:1).

The unit will perform satisfactorily down to at least 8 volts thus giving a good spark at the plugs even under very cold starting conditions when the

battery is at its lowest capacity.

The points and timing should be checked when installing the unit since from now on they can be left for many thousands of miles. Cars fitted with this unit can be expected to operate more economically since the car is kept in "peak tune" all the time. This is because the points do not wear and the timing and dwell angle do not change with time, except for the very slight wear on the heel of the contact breaker.

Four units as described in this article have been built and tested in various types of British car under different driving conditions and it can be stated that "P.E. Scorpio" is without doubt an effective transistor ignition system achieving

improved performance.

Brake horse power has been measured on a "rolling road" and has shown to be about 5 per cent higher with a subsequent improvement in the "liveliness" of the car. One particular feature is the more consistent performance of the car after tuning has occurred from thousands of miles of driving.



#### LOGICAL RADIO CONTROL

continued from page 985

#### **TIMING COMPONENTS**

The interval timing resistors R1 to R5 have a maximum value of approximately 490 ohms due to the input threshold of the gates; lower values may be used with a corresponding change (increase) in the capacitors C2, C4, C6, C8, C10 but the CR product must remain the same.

The signal pulse timing potentiometers VR1 to VR6 will provide a maximum and minimum value. The minimum value of approximately 200 ohms corresponding to a pulse length of approximately 0.5ms and the point at which the following gate will just open. The maximum value is given as 1.2 kilohms. With 10µF capacitors for C3, C5, C7, C9, C11, a pulse of 2ms will be given, while 800 ohms setting for the potentiometers will provide 6ms pulses.

The recommended arrangement is 240\Omega in series with a 500\Omega potentiometer to ground. The cycle time is derived from the trigger circuit (Fig. 5) in which TR1 and TR2 form a multivibrator, the period of which is approximately 1.4CR. If C12 is 1.5\PiF and R6 and R8 are both 15 kilohms, the pulse time will be approximately 18ms (Fig. 6).

The positive going edges at the collector of TR1 are slowed by the action of the timing capacitors, so the output is taken from TR2. Diode D1 and R10 form an isolation circuit so that C13 will not discharge through the trigger circuit. The output waveform is differentiated by C14 and R11 to give a positive pulse of time constant approximately 0.3ms (Fig. 6c). The trigger transistor TR3 switches off producing the sharp-edged negative going pulse (Fig. 6d) to the trigger output every 36ms.

#### LAYOUT

Fig. 7. shows the printed circuit layout full size; it is recommended that extra space along one side is provided for mounting purposes. All holes should be drilled clean and vertical from the copperside with an HSS or tungsten carbide number 58 drill, and a small clean tipped soldering iron used.

Fig. 8 shows component layout, the i.c.s being inserted and soldered in first, followed by the fixed resistors and capacitors. Using thin flexible covered wire the four final interconnections to IC7 should be made as shown in Fig. 9, together with the trigger connections, supply leads and six various coloured wire leads which go to the control panel for connection to VR1 to VR6.

Next month: Two more coders and a decoder

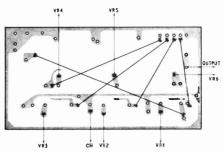


Fig. 9. Potentiometer connections to Coder 1



BY J. D. CROFT

I.C. DIGITAL DICE

"HIS circuit is a revised version of the circuit published in "Ingenuity Unlimited," April 1971, now using integrated circuits.

When re-designing the circuit to use integrated circuits it was found that not only is the finished dice smaller and neater, but by using the low price i.c.s. now available, it is also cheaper to build.

#### **THEORY**

The logic diagram of this dice is shown in Fig. 1.

The circuit is basically a divide-by-six counter driven via the control gate G11 by a 4.8kHz clock

The count is stopped by the control gate when the player presses the play button, and the binary output of the counter is decoded by gates one to nine to light lamps LP1-LP7 set in the usual dice pattern.

The binary output is decoded into four lamp outputs, the first of which decides "even or odd." If the count is odd, the centre lamp, LP3, lights. Next, gate G1 decides "NOT 1" which lights two diagonally opposite lamps LP6 and LP7-except during a "1."

A third gated output, gates G5, G6 and G7 decide 4. 5, or 6 and lights the two remaining diagonally opposite lamps, LP4 and LP5 on these counts. The last output, gates G8 and G9, decide "6" and lights the two remaining lamps on this count. A little thought will show that these combinations will automatically light the correct number of bulbs in the correct patter for each dice position.

Player intervention is prevented in two ways. Firstly, by extinguishing the display lights while the counter is operating, and secondly, by the high clock pulse speed adopted which is far higher than any

human reflex.

#### CIRCUIT CONSTRUCTION

The circuit is built on a piece of Veroboard (0.1in matrix) as shown in Fig. 2. The board should be cut to size and some strips cut as shown in Fig. 3, before the integrated circuits are fitted.

It is recommended to leave the soldering of the i.c.s. until all the other components and straps have been fitted as it is almost impossible to remove them

should they be soldered in the wrong position.

When soldering on 0 lin matrix Veroboard great care should be taken as the solder easily bridges the closely spaced strips, and if the strips are overheated they could come adrift from the laminate causing intermittent faults.

#### COMPONENTS . . .

#### Resistors

RI, R2, R3, R4 4.7kΩ (4 off)

R5 47Ω

R6 100Ω

All 5%, 1W carbon

#### Capacitors

C1, C2 0.1µF 15V (2 off)

Semiconductors TR1, TR2, TR3, TR4 2N3704 (4 off) DI, D2 OA91 (2 off)

#### Integrated circuits

ICI, IC3, IC4 SN7400 (BPOO) Quad 2-input gates

(3 off) IC2 SN7492 (BP92) ÷ 12 counter IC5 SN7410 (BP10) triple 3-input gate

#### Switches

SI Single pole on/off toggleS2 Single pole push betton

LPI-7. 6V 0.3W I.e.s. type (7 off)

#### Miscellaneous

Lampholders for lamps (7 off)

Vercocard,  $2\frac{1}{2}$  in  $\times$   $2\frac{1}{2}$  in (0.1 in matrix) Battery 6V  $(4 \times 1.5$ V in series, type HP7)

#### **CASE**

The dice case may be varied to suit individual requirements and materials available. The case in the suggested design is a standard M.E.M. plastics box for mounting a double mains socket, fitted with the appropriate blanking plate. This will make a neat, strong case, which will accommodate the circuit board and lamp display and also four penlight cells in a four-cell battery holder.

If due to anticipated prolonged use a larger battery or 6V mains power unit is required, a larger case should be used which can be constructed from wood or Perspex.

The lamp display in the author's unit was wired for miniature 6V l.e.s. lamps and lampholders held in position by using stout copper wire, which acts as both support and electrical common positive bus bar. If required the lampholders could be secured in their correct position by using Araldite.

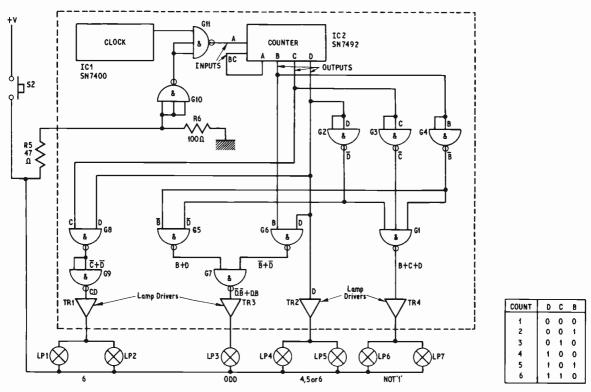


Fig. 1. Circuit diagram of the electronic dice with the logic gates "taken out" of the integrated circuit packs. Circuitry within the grey area is mounted on the 0-lin matrix Veroboard

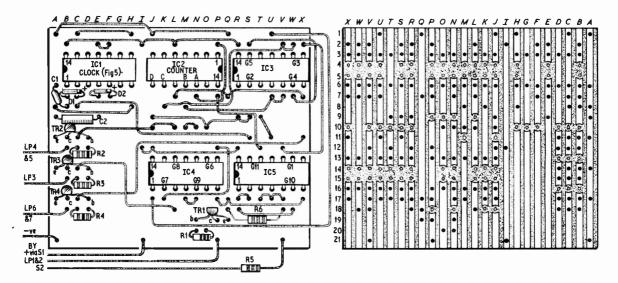


Fig. 2. Wiring and layout of the logic integrated circuits and lamp drivers

Fig. 3. View of the underside of the Veroboard showing the copper regions that must be removed

#### DISPLAY PANEL

The holes for the display are covered by a piece of in translucent Perspex to hide the non-illuminated holes.

This Perspex should be chosen with care, as some colours tend to diffuse the light spots making reading of the display panel difficult.

The Perspex may be selected by holding a sample against a piece of cardboard with a small hole in it and viewing it against background illumination to check that a cleanedged spot can be seen. A blue or violet Perspex was found to be best.

#### CHECKING

After assembling the dice, the connections should be rechecked before fitting the batteries since a single wrong strap could easily permanently damage an integrated circuit.

If desired the correct working of the dice can be verified in the following manner.

#### **CLOCK CHECK**

The clock output can be checked by connecting an oscilloscope between holes 8C and 8A on the Veroboard where a 4.8kHZ square wave should be observed. In the absence of an

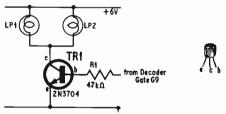


Fig. 4 (a). The lamp driver circuit that amplifies the output from the 4 decoders to light the lamps; (b) Pictorial view of the transistor used in this circuit showing base connections

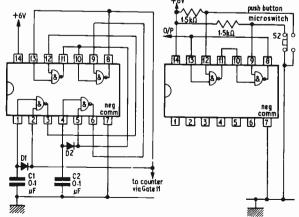
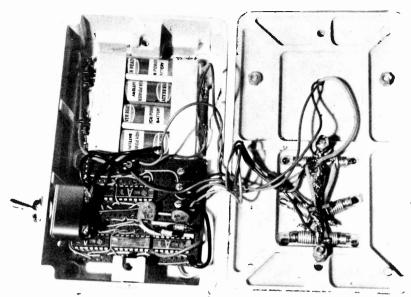


Fig. 5. Wiring of the SN7400N Quad 2-input nand gate to give a 4-8kHz multivibrator for use as clock

Fig. 6. Use of half a SN7400N to provide test input signals to the counter



oscilloscope, a pair of high resistance headphones in series with an  $0.1\mu F$  capacitor may be used to detect the output from the clock which will be observed as a high pitch whistle.

#### TESTING COUNTER AND DECODER

The counter and decoding gates may be checked by injecting test signals into the counter.

This is done by the use of a microswitch buffered by a bistable to remove switching transients. The bistable is constructed from half a SN7400N quad 2-input gate i.c. connected as shown in Fig. 8. This circuit is worth building as a permanent unit as it is invaluable in all testing work on TTL and DTL circuits.

The output from the bistable should be connected to hole 7P on the Veroboard with the wire to 13T temporarily removed. On operating the microswitch the display should be found to advance one count for every two depressions of the microswitch. The reason why two pulses are required to step the display by one count is that the SN7492N is a divide-by-twelve counter, and in the present application the "A" stage output is not used. The reset line of the counter SN7492N is disabled by applying a permanent "O" on pin 7 of the reset gate, which is the Rogate in this package.

#### **SWITCHING TRANSIENTS**

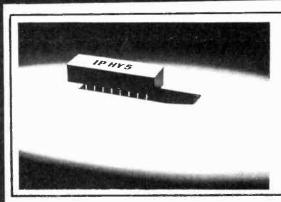
If any trouble is experienced due to an uneven count, noise on the supply line may be the cause, in which case this is rectified by connecting  $0.1\mu F$  capacitors between the negative and positive supply lines as near to the i.c.s as possible. This should remove the sharp fronts of any switching transients transmitted on to the supply lines.

This expedient should not be required if the unit is powered by batteries but may be required if the unit incorporates a mains power supply.

As the i.c.s are already working at 0.75 volts over the manufacturers' recommended working voltage, on no account must the supply voltage be raised above six volts.

Providing this value is not exceeded, no damage to the i.c.s should occur.

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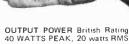
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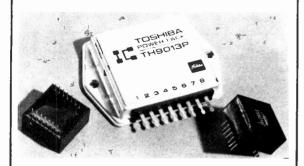
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185	-22	30C15	-58	DY802	-35	EM80	-41	PCL83	-57	UAF42	-51
1T4	-16	30C17	.79	EABC80	-32	EM81	-41	PCL84	-34	UBC41	-52
384	-26		-61	EAF42	-50	EM84	-32		-38	UBF80	-84
3 V 4	-37	30F5	-69	EB41	-40	EM87	-36	PCL86	-39	UBF89	-32
5U4G	-26	30FL1	-61	EB91	-11	E¥51	-33		•65		-33
5V4G	-35	30FL12	•70	EBC33	-40	EY86	-29				-35
5Y3GT	-26	30FL14	-68	EBC41	-54	EZ40	-43		-77		-88
5Z4G	.85	30L1	-29	EBC90	-22	EZ41	·43	PEN360		UCH42	-58
6/30L2	- 54	30L15	-57	EBF80	-32	EZ80	-22				-82
6AL5	-11	30L17	-71	EBF89	-29	EZ81	.23	PL36	-49	UCL82	-32
6AM6	-13	30 P4	-57	ECC81	-17	GZ30	-35	PL81	-44	UCL83	-55
6AQ5	-22	30P12	.72	ECC82	-20	GZ32	.40	PL81A	-49		-56
6AT6	.20	30P19	-57	ECC83	∙35	GZ34	-48	PL82	-81	UF89	-80
6AU6	-20	30PL1	-63	ECC85	-26	KT41	-77		-88	UL41	-57
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6BJ6	-41	30PL15	-90	ECF82	-26	LN319	-63	PL504	-63	UM84	-22
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6F23	-68	807	-45	ECH83	-40	P61	-45	PY33	-55	Z77	-22
6F25	.57	6063	-62	ECH84	-36	PABC80	.34	PY81	-25	Transis	tors
6K7G	-12	AC/VP2	.77	ECL80	-30	PC86	-47	PY82	-25	AC107	.17
6K8G	-17	B349	-65	ECL82	-31	PC88	-47	PY83	-28	AC127	-18
6Q7G	.27	B729	-82	ECL86	-36	PC96	-42	PY88	-83	AD140	-37
68N7GT	-30	CCH35	-67	EF39	-38	PC97	-39	PY800	-34	AF115	-20
6V6G	-23	CY31	•80	EF41	-60	PC900	.33	PY801		AF116	-20
6V6GT	-31	DAF91	.22	EF80	-23	PCC84	-29	R19	-30	AF117	-20
6 X 4	-23	DAF96	-36	EF85	-28	PCC85	-27	R20	-56	AF118	-48
6X5GT	.28	DF33	-38	EF86	-30	PCC88	.42	U25	-64	AF125	.17
7B7	.38	DF91	.16	EF89	-26	PCC89	•45	U26	-56	AF127	-17
10P13	-58	DF96	-36	EF91	.13	PCC189	-48	U47		OC26	-25
12AT7	-17	DH77	-20	EF98	-65	PCC805	. 56	U49		OC44	-12
12AU6	-20	DK 32	-33	EF183	-28	PCF80		U50		OC45	-12
12AU7	-20	DK91	-28	EF184	-31	PCF82		U52		OC71	-12
12AX7	.22	DK92	-38	EH90	-37	PCF86		U78		OC72	-12
19BG6G	-87	DK96	-38	EL33	-55	PCF800		U191		OC75	-12
20F2	-67	DL35	-40	EL34	.45	PCF801		U193		OC81	-12
20P3	-80	DL92 DL94	-26	EL41 EL84	-54	PCF802		U251		OC81D	-12
20P4	-92		-37		.23	PCF805		U301		OC82	-12
25L6GT	-20	DL96	-38	EL90	-26	PCF806		U329		OC82D	-12
25U4GT	-57	DY86	-25	EL95	·33 '	PCF808	-68	108U	·98 '	OC170	-22

#### READERS RADIO

85 TORQUAY GARDENS, REDBRIDGE, ILFORD, ESSEX. Tel. 01-550 7441.

Postage on 1 valve 5p, on 2 or more valves 3p per valve extra.

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A look at most of the linear and logic integrated circuits now readily available to experimenters, private More detailed technical constructors and designers. information and pin connection diagrams are usually to be found in manufacturers' literature, often available from i.c. suppliers. Devices shown as being in TO5 packages may also be in similar packages (e.g. TO99, TO100, etc.) but with a different number of lead-outs.

PART 1 LINEAR I.Cs

"HE first linear integrated circuits to be produced were operational amplifiers designed for instrumentation and similar applications. Introduced by Fairchild in 1963 the nA702 gave a performance equal to or better than the available discrete component amplifiers at a similar cost. The first "high performance" operational amplifier was the #A709 which Fairchild introduced in 1965. It had more gain, a reduction of input offset current, and allowed a higher output level.

While all operational amplifiers have many characteristics in common, some conditions peculiar to individual amplifiers may be found. For example, the #A702 has a limited input voltage range of  $\pm 1.5$ V to  $\pm 6.0$ V indicating the possibility of "latch-up"

is exceeded

**Circuit Function** 

amplifier 7020

amplifier 709C Dual version of 709C in one

performance operational amplifier Version 741C without compensation 748C

Dual version of 741C

package

High speed differential comparator 710C

General purpose operational

High performance operational

Frequency compensated high

The #A709 output stage uses complementary emitter followers with no bias between the two bases. This produces a dead zone which can cause an excessive distortion of the output signal when only a small amount of overall negative feedback is used.

The first of the "second generation" amplifiers was the #A741. This has built-in frequency compensation, short-circuit protection, no crossover distortion, an offset voltage null capability, low power consumption, no latch-up, and a large common-mode and differential voltage range.

At first sight this built-in solve the problems of operat without instability occurring ponents. However, it means NOT optimum for overall gain

For a 40dB gain the 741 with the 1MHz bandwidth of mended compensation.

\*Packag

DIL

T05

DIL

T05

DIL

T05

T05

DIL

tir t ns	ng down and wit hat the greater as a bar	to an or hout us frequence than unitidwidth	verall g sing ex cy comp ty. of 10kH	appears ain of un ternal co pensation	red	4			
00	namable	with th	e 709	and rec	om-		1		
ge	Ferranti	Motorola	Mullard	National Semi- conductor	RCA	RS Com- ponents	Siemens	sgs	T∉xas
		MC1710CG MC1710CP		LM710C LM710CN		710MOPA 710 OPA		U5B771039X U6E7710393	SN72710L SN72710N
		MC1712CG MC1712CP	TAA241		CA3032			U5B771239X U6E7712393	SN72702L SN72702N
		MC1709CG MC1709CP		LM709C LM709CN		709MOPA 709OPA	TA A 25 I	U5B770939X U6E7709393	SN72709L SN72709N
		MC1437P							SN72709DN
		MC1741CG MC1741CP		LM741C LM741CN	CA3056	7410PA	TBA22I	L141T1 L141B1	SNT2741L SNT2741N
			TAASII	LM748C					SN72748L
				LM747C				L147B1	SN72747N

DIL packages here are 14-lead versions except 7410PA which has 8 leads. Texas also make SN72709P. SN72741P. SN72748P in 8-lead DIL Most manufacturers make operational amplifiers and comparators

#### **FERRANTI** GENERAL ELECTRIC A.C. AMPLIFIERS A.C. AMPLIFIERS TYPE **FUNCTION** CASE **TYPE FUNCTION** ZLA10 Wideband linear amplifier for video and Old Number **New Number** i.f. stages T05 PA222 GEL222FI 800mW audio amplifier. ZLA15 Wideband linear amplifier, improved $22\Omega$ load version of ZLA10 T05 PA230 GEL230FI Low level audio preamplifier. 72db gain PA234 **GEL234FI** I W audio amplifier. $22\Omega$ load D.C. AMPLIFIERS PA237 **GEL237FI** 2W audio amplifier. 16 $\Omega$ load ZLD2S/T/U Differential d.c. amplifiers T05 PA239 GEL239FI Stereo version of PA230 ZLD12C/D Complementary output stages for ZLD2 PA245 GEL246SI 5W audio amplifier, $16\Omega$ load T05 series PA263 GEL263S1 3.5W audio amplifier ZLD7090 Operational amplifier T05 PA266 GEL266FI I 6W audio amplifier ZLD709CE Operational amplifier DIL ZLD741C Operational amplifier **T05** OTHER CIRCUITS ZLD741CE Operational amplifier DIL TYPE TYPE **FUNCTION** PA264 Voltage regulator **GEL264SI** PA265 GEL265\$1 OTHER CIRCUITS Voltage regulator PA424 GEL300FI Zero voltage switch ZN400E Analogue switch PA436 GEL301FI DIL Phase controller for SCR ZN402E Gated operational amplifier DIL PA494 GEL304A1 Threshold detector ZN403E Servo amplifier DIL All the above devices have a special package MOTOROLA

	IVIC		RULA		
	A.C. AMPLIFIERS			D.C. AMPLIFIERS	
TYPE	FUNCTION	CASE	TYPE	FUNCTION	CASE
MC1303L	Stereo pre-amplifier. 68dB gain.		MC1430P	Operational amplifier 69dB gain	DIL
	60dB separation	DIL	MC1433G	Operational amplifier 89dB gain	T05
MC1304P	Stereo demodulator for f.m.	DIL	MC1435G	Dual operational amplifiers of 71dB	
MC1305P	Stereo demodulator for f.m.	DIL		gain	T05
MC1306P	Complementary amplifier and pre-		MC1435L	Dual operational amplifiers	DIL
	amplifier	DIL	MC1435P	Dual operational amplifiers	DIL
MC1307P	Stereo demodulator for f.m. No	5,2	MC1437L	Dual MC1709	DIL
	audio mute	DIL	MC1439G	Internally compensated operational	
MC1330P	Low level video detector	DIL		amplifier of 84dB gain	T05
MC1350P	I.F. amplifier with a.g.c.	DIL	MC1709CP	Operational amplifier	DIL
MC1445L	Wideband amplifier circuits 16dB	DIL	MC1710CP	Differential comparator	DIL
	qain	DIL	MC1712CP	Operational amplifier	DIL
MFC6010	F.M./I.F. differential amplifier	5.2	MC1741CP	Fully compensated op. amplifier	DIL
	40dB gain at IOMHz	S		OTHER CIRCUITS	
MFC8000	Dual differential amplifiers de-			OTHER CIRCUITS	
MFC8001	signed for input stage of stereo	S S	TYPE	FUNCTION	CASE
MFC8002	power amplifiers	Š	MC1460G	Voltage regulator 2.5 to 17V	
MFC8010	I W audio amplifier and preamp	Š	111011000	250mA	T05
MFC8020	Class B driver	Š	MC1461G	Voltage regulator 2.5 to 32V	
MFC8030	Flexible differential amplifier, 32dB	•		250m A	T05
	gain	S	MC1463G	Voltage regulator 3.8 to 32V	
MFC8040	Audio preamp 3dB noise, 80dB			250mA	T05
	gain	S	MC1466L	Wide range voltage and current	
MFC9000	4W audio amplifier and preamp	Š		regulator	DIL
MFC9010	2W audio amplifier and preamp	Š	MC14S9R	Voltage regulator 2.5 to 32 volts	
	,	•		600mA	*
0 0			MFC4060	Regulator 4V 200mA	S
	package		M FC6030	Regulator 4V 200m A	Š
" 9-lead sm	all power transistor case		MFC6040	Electronic attenuator	S S

### NATIONAL SEMICONDUCTOR & SILICON GENERAL

D.0	C. AMPLIFIERS		0	THER CIRCUITS	
TYPE	FUNCTION	CASE	TYPE	FUNCTION	CASE
LM709CN/SG709CN	Operational amplifier	DIL	LM103*	Voltage regulator I-8V to	
LM710CN/SG710CN	Differential comparator	DIL		5.6V available	T046
LM7f1CN/SG711CN	Core memory sense amplifier	DIL	LM305/SG305T	Positive voltage regulator	
LM741C/SG741CT	Fully compensated opera-			4-5 to 30 V 20m A	T05
	tional amplifier	T05	LM309/SG309T	Voltage regulator 5V 200 mA.	
LM748C/SG748CT	As LM741C without com-			No external components	T05
	pensation	T05	LM309K/SG309K	Voltage regulator 5V IA.	
	·			No external components	T03
	*LM103 is specified as "LM	1031-8	" etc. depending on	voltage required.	

	N	/UI	LARD		
	A.C. AMPLIFIERS			D.C. AMPLIFIERS	
TYPE	FUNCTION	CASE	TYPE	FUNCTION	CASE
TAA263	Three stage cascade amplifier	T072	TAA241	Operational amplifier (702C)	T05
TAA293	Medium frequency, general pur-		TAA242	Operational amplifier (702)	T05
.,	pose amplifier	T074	TAA521	Operational amplifier (709C)	T05
TAA300	Audio amplifier. I watt into 80hms	T074	TAA811	Operational amplifier (741C with-	TOF
TAA310	Low noise audio preamplifier for	T074	TD 4 00 1	out compensation) Fully compensated operational	T05
T A A 200	tape recorders  Metal oxide silicon l.f. preamplifier	T074 T018	TBA221	Fully compensated operational amplifier (741C)	T05
TAA320 TAA350	Wideband limiting f.m./l.f. ampli-	1010	TBA222	Fully compensated operational	
1 A A 3 3 U	fier	T074	IDALLE	amplifier (741)	T05
TAA570	Limiter-amplifier with f.m				
	detector	T05			
TAA960	Three stage active filter amplifier.				
	39dB gain per stage	T074		OTHER CIRCUITS	
TAA970	Microphone amplifier Four transistor ring modulator/	T074	TYPE	FUNCTION	CASE
TAB101	demodulator	T074	TBA281	Voltage regulator 2 to 37V 150mA	<b>T</b> 074
TAD100	A.M. receiver circuit	DIL	TBA673	Modulator/demodulator. Ring of	
TADIIO	A.M./F.M. i.f. amplifiers, three			four transistors	T074
,,,,,,,,,	gain blocks	DIL	TBA750	Limiter amplifier	16DIL
		A.C. AN	SSEY IPLIFIERS		
TYPE	FUNCTION	CASE	\$L621C	SSB a.g.c. generator used with	T05
SL201B	Video amplifier, current gain 26, 15MHz BW	T05	SL630C	\$L610/11/12 Microphone/headphone amplifier	
\$L301B	Matched pair of transistors, hee			with a.g.c. Gain 40dB	T05
	25 to 250, f <sub>T</sub> 600MHz	T05	SL640C	Double balanced modulator	T05
\$L303B	Matched trio of transistors, her	T05	\$L641C	Receiver mixer, double balanced modulator	T05
01.400.0	25 to 250, f <sub>T</sub> 600MHz	T05			103
<b>SL</b> 403D	3 watt audio amplifier 8 ohms load, 18V supply	*	01.7010/0	D.C. AMPLIFIERS	
SL610C	R.F./I.F. amplifiers. 18dB gain at		SL701B/C	Operational amplifier, output sym- metrical about earth	T05
320100	30MHz. 4dB noise. 85MHz BW	T05	SL702B/C	Operational amplifier, output not	. 03
SL611C	R.F./I.F. amplifiers. 24dB gain at		32102070	symmetrical. The minimum gain	
	30MHz. 4dB noise. 50MHz BW	T05		is 66dB, open loop bandwidth	
SL612C	R.F./I.F. amplifiers. 32dB gain at	TOF		250kHz on all \$L701, 702 i.c.s	T05
01 0000	1-7MHz. 3dB noise. IOMHz BW	T05		TEXAS	
\$L620C	Voice operated gain adjusting device used with SL630C	T05	Teyas Ins	struments produce all the common ope	erational
* Consist =s	ackage with heat sink, DIL style	. 00	amplifier cir	rcuits (702, etc.)	
SL201 B/C	are now maintenance types		SN76013N 4	4W audio amplifier similar to Sinclair	r 1012
		R.	C.A.		
	A.C. A		. AMPLIFIE	RS	
TYPE	FUNCTION	CASE	TYPE	FUNCTION	CASE
CA3000	Differential amplifier gain 28dB		CA3019	Package with six diodes	T05
37,000	BW 650kHz	T05	CA3020	Wideband amplifier gain 75dB BW	
CA3001	Differential amplifier gain 16dB			8MHz	T05
	BW 16MHz	T05	CA3021	Wideband amplifier gain 50dB BW	T05
CA3002	R.F./I.F. amplifier gain 19dB		C A 2000	800kHz	T05
0.4.2005	BW IIMHz	T05	CA3022	Wideband amplifier gain 50dB BW 3MHz	T05
CA3005	R.F./I.F. differential amplifier gain		0.4.0000	Middhaud amulifian nain FOdD DW	. 03

	A.C. AN	D.C.	AMPLIFIE	:RS	
TYPE	FUNCTION	CASE	TYPE	FUNCTION	CASE
CA3000	Differential amplifier gain 28dB		CA3019	Package with six diodes	T05
	BW 650kHz	T05	CA3020	Wideband amplifier gain 75dB BW	
CA3001	Differential amplifier gain 16dB		0.4.0001	8MHz	T05
0.4.2000	BW 16MHz	T05	CA3021	Wideband amplifier gain 50dB BW 800kHz	T05
CA3002	R.F./I.F. amplifier gain 19dB BW 11MHz	T05	CA3022	Wideband amplifier gain 50dB BW	.03
CA3005	R.F./I.F. differential amplifier gain	103	07100==	3MHz	T05
0710000	16dB BW 120MHz	T05	CA3023	Wideband amplifier gain 50dB BW	
CA3007	Differential audio amplifier gain			_ IOMHz	T05
	20dB BW 20kHz	T05	CA3026	Two darlington transistor pairs	T05
CA3010	Operational amplifier gain 57dB	T05	CA3028A	$h_{FE}$ of 110 $f_T$ of 550MHz R.F./I.F. differential amplifier gain	103
CA3011	BW 200kHz Wideband amplifier gain 65dB BW	105	UMSUZOM	35dB BW 5MHz	T099
UMSUTT	20MHz	T05	CA3029	Operational amplifier gain 57dB	
CA3012	Wideband amplifier gain 65dB BW			BW 200kHz	DIL
	20 MHz	T05	CA3030	Operational amplifier gain 66dB	
CA3013	Wideband amplifier gain 65dB BW		0.4.0005	BW 200kHz	DIL
0.4.0014	20 MHz	T05	CA3035	Wideband amplifier gain 40dB BW 2.5MHz	T05
CA3014	Wideband amplifier gain 65dB BW 20MHz	T05	CA3036	Two darlington pair transistors	
CA3015	Operational amplifier gain 66dB	103	0710000	h <sub>FE</sub> 82 f <sub>T</sub> 150MHz	T05
0710010	BW 200kHz	T05	CA3039	Six diodes 5V p.i.v. 25mA max.	
CA3018	Package with transistors hee of 30			current	T05
	f <sub>™</sub> 300MHz	T05			

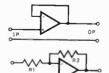
			R.C.	A.		-80-
TYP	PΕ	FUNCTION	CASE	TYPE	FUNCTION	CASE
CA3	040	Wideband amplifier gain 34dB BW		CA3055	Voltage regulator I-8 to 34V II5mA	T05
		40MHz	T05	CA3056	Operational amplifier gain 86dB	
CA3	104 I	Wideband amplifier/f.m. detector/			(741C)	T05
	10.40	a.f. preamp	14 <b>Q</b>	CA3059	Zero voltage switch for thyristor	
CA3	1042	Wideband amplifier/f.m. detector/ a.f. preamp	140	C & 20C0	control 105mA	DIL
CAS	1043	1.F. amplifier/limiter/f.m. detector/	140	CA3060	Operational amplifier, characteris- tics can be varied by user	DIL
OAG	,010	audio preamp	T05	CA3062	Photo detector and power ampli-	DIL
CA3	044	Wideband amplifier/phase detector,			fier	T05
		for a.f.c. systems	T05	CA3065	I.F. amplifier, limiter, f.m. detec-	
		Five transistors her 110 ft 300MHz	DIL		tor, audio driver	140
		Five transistors her 110 ft 300MHz	DIL	CA3070	Television chroma system, chroma	
CA3	047	Operational amplifier gain 84dB BW 20kHz	DIL	CA3071	signal processor	DIL
CA3	nas	Audio amplifier 53dB gain BW	DIL	CA30/1	Television chroma system, chroma amplifier	DIL
UAU		250kHz	DIL	CA3072	Television chroma system, chroma	DIL
CA3	1049	Six npn transistors fr 1.3GHz	T05	01.001.2	demodulator	DIL
CA3	050	Two darlington differential ampli-		CA3075	F.M./I.F. amplifier, limiter, detec-	
		fiers f <sub>T</sub> 600MHz	DIL		tor, audio preamplifier	140
CA3	1051	Two darlington differential ampli-	DIL	CA3078	Operational amplifier gain 80dB	
		fiers f <sub>T</sub> 600MHz			BW 20MHz	T05
CV3	052		5.2			
CA3	052	Audio amplifier gain 53dB BW	DIL	Most of	the values given for RCA i.c.s are m	inimum
CA3					the values given for RCA i.c.s are meet that in some cases the bandwidth fig	
CA3	1053	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz	DIL T099	values, exc wideband (	ept that in some cases the bandwidth fig devices are the maximum useful freque	ures for
	1053	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain	DIL	values, exc wideband (	ept that in some cases the bandwidth fig	ures for
CA3	053   054	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB	DIL T099	values, exc wideband o the 3dB do	ept that in some cases the bandwidth fig devices are the maximum useful freque own frequency	ures for ncy, net
CA3	5.C	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB	DIL TO99 DIL	values, exc wideband of the 3dB do	cept that in some cases the bandwidth fig devices are the maximum useful frequent wn frequency EMENS A.C. AMPLIFIE	ures for ncy, not RS
CA3	5.C	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB S.S. A.C. AMPLIFIERS FUNCTION	DIL T099 DIL	values, exc wideband of the 3dB do	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE FUNCTION	ures for ncy, net
CA3	5.C PE A611B	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier	DIL T099 DIL CASE	values, exc wideband of the 3dB do	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB	res for ncy, net
CA3 CA3 TYE	5.053   1054   1	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB S.S. A.C. AMPLIFIERS FUNCTION	DIL T099 DIL CASE S	values, exc wideband of the 3dB do SIE TYPE TAAI4I	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW	ncy, net
CA3 CA3 TYP	5.053 S.C.PE A611B A611C	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier	DIL T099 DIL CASE	values, exc wideband of the 3dB do	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB  gain 20kHz BW  Three stage audio amplifier 70dB  gain 600kHz BW	RS CASE
CA3 CA3 TYF TAA TAA TAA TAA	5.053 5.054 PE A611B A611C A621 A621A A661B	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier F.M./I.F. amplifier and detector	DIL T099 DIL CASE S	values, exc wideband of the 3dB do SIE TYPE TAAI4I	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB  gain 20kHz BW  Three stage audio amplifier 70dB  gain 600kHz BW	res for ncy, net
CA3 CA3 TYF TAA TAA TAA TAA	5.053 5.054 PE A611B A611C A621 A621A A661B	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  Color of the state of the s	DIL TO99 DIL CASE S S S S	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151	cept that in some cases the bandwidth fig devices are the maximum useful frequency in frequency.  EMENS A.C. AMPLIFIE FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW	RS CASE T072 T0100
CA3  CA3  TYF  TAA  TAA  TAA  TBA	5.053 5.054 S.C PE A611B A611C A621 A621A A661B A231	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  Color of the second o	DIL T099 DIL CASE S S S S DIL	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435	cept that in some cases the bandwidth fig devices are the maximum useful frequency in frequency  EMENS A.C. AMPLIFIE FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages	RS CASE T072
CA3 CA3 TYP TAA TAA TAA TAA TBA	0053   1054   1   1054   1   1054   1   1054   1   1   1   1   1   1   1   1   1	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier 5.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode	DIL T099 DIL CASE S S S S DIL T018	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151	cept that in some cases the bandwidth fig devices are the maximum useful frequency.  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and de-	RS CASE T0100 T0100 T0100
CA3 CA3 TYP TAA TAA TAA TAA TAA TBA	0053   1054   1   1054   1   1054   1   1054   1   1   1   1   1   1   1   1   1	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier 5.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing	DIL TO99 DIL  CASE S S S S DIL TO18 DIL	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain	RS CASE T072 T0100
CA3 CA3 TYF TAA TAA TAA TAA TBA TBA	5.053   1054   1	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier 5.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode	DIL T099 DIL  CASE S S S S DIL T018	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain  Broadband amplifier r.f./i.f. 75dB	TO100 TO100 DIL
CA3  TYP  TAA  TAA  TAA  TAA  TAA  TBA  TBA  TB	0053 0054 S.C PE A611B A611C A621A A661B A231 A271 A311 A318 A591 A641A	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  C.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier 4W audio amplifier fier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier, detector, preamp 2.5W audio amplifier	DIL TO99 DIL CASE S S S S DIL TO18 DIL S S S	values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain	RS CASE TO72 TO100 TO100 DIL S
CA3  CA3  TYF  TAA  TAA  TAA  TBA  TBA  TBA  TBA  TB	0053 0054 1054	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier f.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier, detector, preamp 2.5W audio amplifier 4W audio amplifier	DIL TO99 DIL CASE S S S S DIL TO18 DIL S S S S S S S S S S S S S S S S S S S	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400	cept that in some cases the bandwidth fig devices are the maximum useful frequent with frequent frequency.  EMENS A.C. AMPLIFIE FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain  Broadband amplifier r.f./i.f. 75dB power gain at 35MHz	TO100 TO100 DIL
CA3 CA3 TYF TAA TAA TAA TAA TBA TBA TBA TBA TBB TBB	5.053 1054 PE AGIIB AGIIC AG2IA AG2IA AG2IA AG2IA AG2IA AG2IA AG2IA AG4IA AG4IA AG4IA AG4IA AG4IA AG4IB	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier F.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier, detector, preamp 2.5W audio amplifier 4W audio amplifier A.M. receiver circuit	DIL TO99 DIL  CASE S S S S DIL TO18 S S S S S S S S S S S S S S S S S S S	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400 TBA450	cept that in some cases the bandwidth fig devices are the maximum useful frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain  Broadband amplifier r.f./i.f. 75dB power gain at 35MHz  Stereo decoder  A.M./F.M. i.f. 1.f. amplifier	RS CASE TO72 TO100 TO100 DIL S
CA3 CA3 TYF TAA TAA TAA TAA TBA TBA TBA TBA TBB TBB	1053   1054   1   1054   1   1054   1   1054   1   1054   1   1   1   1   1   1   1   1   1	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier F.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier, detector, preamp 2.5W audio amplifier A.M. receiver circuit Television sound circuit	DIL TO99 DIL CASE S S S S DIL TO18 DIL S S S S S S S S S S S S S S S S S S S	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400 TBA450 TBA460	cept that in some cases the bandwidth fig devices are the maximum useful frequent own frequency  EMENS A.C. AMPLIFIE FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain Broadband amplifier r.f./i.f. 75dB power gain at 35MHz  Stereo decoder A.M./F.M. i.f. 1.f. amplifier  D.C. AMPLIFIERS	TO100 TO100 DIL S S
CA3  CA3  TYF  TAA  TAA  TAA  TBA  TBA  TBA  TBA  TB	0053 0054 S.C PE AGIIB AGIIC AG2IA AG2IA AG2IA AG2IA AG4IB AG4IA AG4IA AG4IB AG6IB	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  C.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 5.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier 4W audio amplifier 4W audio amplifier 4W audio amplifier A.M. receiver circuit Television sound circuit D.C. AMPLIFIERS	TO99 DIL  CASE S S S S DIL TO18 S S S S S S S S S S S S S S S S S S S	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400 TBA460	cept that in some cases the bandwidth fig devices are the maximum useful frequent own frequency  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Three stage audio amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain  Broadband amplifier r.f./i.f. 75dB power gain at 35MHz  Stereo decoder  A.M./F.M. i.f. I.f. amplifier  D.C. AMPLIFIERS  FUNCTION	TO100 TO100 DIL S S CASE
CA3  TYF  TAA  TAA  TAA  TAA  TBA  TBA  TBA  TB	0053 0054 S.C PE A611B A611C A621A A621A A621A A621B A271 A3311 A3311 A581 A581 A641A A641B A661B A661B	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  C.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier F.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing I.F./F.M. amplifier, detector, preamp I.F./F.M. amplifier 4W audio amplifier 4W audio amplifier A.M. receiver circuit Television sound circuit D.C. AMPLIFIERS High speed differential comparator	DIL T099 DIL CASE S S S S DIL T018 DIL S S S S T05	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400 TBA460 TBA460	cept that in some cases the bandwidth fig devices are the maximum useful frequent of the property of the prope	TO100 TO100 TO100 DIL S S CASE TO5
CA3  TYP  TAA  TAA  TAA  TAA  TAA  TBA  TBA  TB	0053 0054 1054 1054 1054 1054 1054 1054 1054 1054 1055	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  S.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier 4W audio amplifier 4W audio amplifier 4W audio amplifier 5.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing 1.F./F.M. amplifier, detector, preamp 1.F./F.M. amplifier, detector, preamp 2.5W audio amplifier 4W audio amplifier A.M. receiver circuit Television sound circuit D.C. AMPLIFIERS High speed differential comparator General purpose operational amplifier	DIL T099 DIL CASE S S S S DIL T018 DIL S S S S T05	values, exc wideband of the 3dB do  SIE  TYPE  TAA141  TAA420  TAA435  TBA420  TBA400  TBA460  TYPE  TAA521  TAA522	cept that in some cases the bandwidth fig devices are the maximum useful frequential frequential frequential frequential frequential frequential frequential frequency.  EMENS A.C. AMPLIFIE  FUNCTION  Three stage audio amplifier 70dB gain 20kHz BW  Low noise five stage amplifier 70dB gain 600kHz BW  Low noise five stage amplifier 70dB gain 20kHz BW  Audio amplifier for driver stages F.M./I.F. amplifier and demodulator 60dB gain  Broadband amplifier r.f./i.f. 75dB power gain at 35MHz  Stereo decoder  A.M./F.M. i.f. 1.f. amplifier  D.C. AMPLIFIERS  FUNCTION  Operational amplifier (709C)  Operational amplifier (709C)	RS CASE T072 T0100 T0100 DIL S S CASE
CA3  TYF  TAA  TAA  TAA  TAA  TBA  TBA  TBA  TB	1053 1054	Audio amplifier gain 53dB BW 300kHz R.F./I.F. differential amplifier gain 35dB BW 5MHz Differential amplifier gain 32dB  C.S. A.C. AMPLIFIERS FUNCTION 2W audio amplifier 3W audio amplifier 3W audio amplifier 4W audio amplifier F.M./I.F. amplifier and detector Dual low noise operational amplifier gain 70dB Voltage stabiliser for varicap diode Television signal processing I.F./F.M. amplifier, detector, preamp I.F./F.M. amplifier 4W audio amplifier 4W audio amplifier A.M. receiver circuit Television sound circuit D.C. AMPLIFIERS High speed differential comparator	DIL T099 DIL  CASE S S S S DIL T018 DIL S S S S T05 T05	Values, exc wideband of the 3dB do SIE TYPE TAA141 TAA151 TAA420 TAA435 TBA120 TBA400 TBA460 TBA460	cept that in some cases the bandwidth fig devices are the maximum useful frequent of the property of the prope	TO100 TO100 TO100 DIL S S CASE TO5

Both the 709 and 702 require several external components to frequency-compensate the amplifier so that it is stable under closed-loop conditions and cannot oscillate. The 702, 709 and the 710 comparator were the "first generation" amplifiers.

Although fully compensated, operational amplifiers can be used in any feedback circuit without instability occurring, and without external compensation components, bandwidth and slew rate are reduced. The 710, 702, 709, 741 are produced by many manufacturers. A "C" suffix (for example 702C) has a slightly reduced temperature range and specification over the version without the Suffix

#### SERIES 741C OPERATIONAL AMPLIFIER

Although operational amplifiers are perhaps the most useful of the vast range of available linear integrated circuits, with varied applications from multivibrators to active filters, many circuits are designed for specific applications. A glance at this survey will reveal a small part of the range available to commercial and industrial organisations. Most of those listed are also readily available to the home constructor.

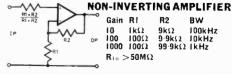


#### **UNITY GAIN VOLTAGE FOLLOWER**

Input impedance Output impedance  $1\Omega$ Bandwidth **IMHz** 

#### INVERTING AMPLIFIER Gain R1 R2 BW

 $10k\Omega$   $10k\Omega$ IMHz IkΩ IkΩ 10kΩ 100kΩ 100kHz 100 1kΩ 100kΩ 1000 100Ω 100kΩ 10kHz  $R_{\rm lit} < 10 k\Omega$ 



~	Gain	RI	R2	BW
	10	lkΩ	9kΩ	100kHz
)P	100	1000	9-9k(2	JOKHZ
	1000	1000	99-9kΩ	IkHz
	Rin >	>50MΩ		
-				

Logic integrated circuits have opened up a whole new science in systems design that does not necessitate deep technical knowledge of the circuit inside the package. Given a basic set of parameters, logic i.c.s can be built into large systems with little difficulty.

REPRESENTING almost unbelievable value for money in terms of components per penny (the 7400 TTL gate, for example, can be purchased for as little as 15p and contains sixteen transistors, sixteen resistors and four diodes), the logic i.c. families are now available to the amateur from many suppliers who stock an incredibly wide range of types.

To those readers put off by such mystic terms as "fan-out", "flip-flop" and "truth-table", take heart, almost anything that can be built with logic i.c.s can also be built (mentally at least) using relays. Logic gates are simply switches with an "on" and

"off" state, and all the more complicated devices, such as bistables, can be built with gates. Logic really is easy to pick up and after a start has been made, the subtleties can be appreciated one by one.

For those who have always wanted to design something electronic, but have been put off by simultaneous equations and slide rules, this is the scene for you; after a comprehension of the principles is obtained, all you need to start designing is common sense; all the maths has been done by the manufacturers.

#### LOGIC FAMILIES

The three main families are RTL, DTL, and TTL, while MSI (medium scale integration) uses TTL methods and incorporates several TTL circuits in each package to give a complex function, for example, a counter.

The tables given here are far from exhaustive because of the several manufacturers' different type numbers, but they do represent the majority of devices available through retail outlets.

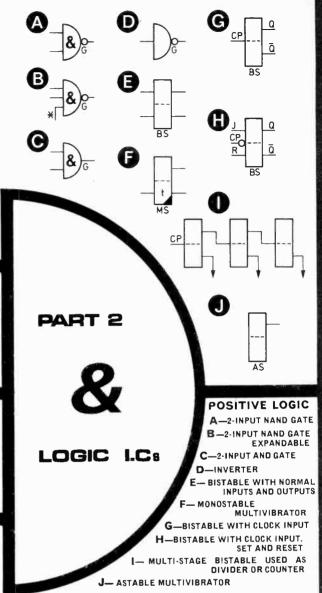
RTL is gradually losing favour because of the competitive lowering of prices of the DTL and TTL.

In the DTL range, although a common coding can be found among many manufacturers, it is also worth obtaining the leaflets and handbooks on other types. Mullard and Siemens operate Pro Electron codings but they are not always interchangeable equivalents

in all cases. Ferranti use different codings again for their Micronor series, prefixed ZS, which is divided into a 20ns range, medium speed range (15ns), and a high speed range (9ns).

Readily available types in the DTL range are mostly given a coding which includes digits 930 and 9090 upwards, other letters and figures being added according to manufacturers' choice. The Mullard range uses Pro Electron numbering prefixed by FC.

The widest range is TTL which extends into medium scale integration MSI. Again because of space limitations we have listed the most common in the 74 series, various manufacturers adding their own personal identity prefix. It is worth pointing out here that although many of the Mullard and Siemens TTL types (prefix FJ) are similar to the 74 series, they are not all identical.



FAIRCHIL	D FERRANTI	ITT	MOTOROLA	MULLARD	NATIONAL	SIEMENS	TEXA8
9002	ZN7400E	9002	MC7400P	FJH131	DM8000N	FLHIOI	SN7400N
****	ZN7401E		MC7401P	FJH231	DM8001N	FLH201	SN7401N
	ZN7402E		MC7402P	FJH221	DM8002N	FLH191	SN7402N
				FJH291	DM8003N		SN7403N
	ZN7404E	9016	MC7404P	FJH241	DM8004N		SN7404N
9003	ZN7410E	9003	MC7410P	FJH121	DM8010N	FLHIII	SN7410N
9004	ZN7420E	9004	MC7420P	FJHIII	DM8020N	FLH121	SN7420N
	ZN7430E	•	MC7430P	FJHIOI	DM8030N	FLH131	SN7430N
9009	ZN7440E	9009	MC7440P	FJH141	DM8040N	FLHIGI	SN7440N
0000	ZN7441E			FJLI01	DM8840N		SN7441A
				FJH261	DM8842N		SN7442N
9005	ZN7450E	9005	MC7450P	FJH151	DM8050N	FLH151	SN7450N
0000	ZN7451E		MC7451P	FJH161	DM8051N	FLH161	SN7451N
	ZN7453E	9008	MC7453P	FJH171	DM8053N	FLH171	SN7453N
	ZN7454E		MC7454P	FJH181	DM8054N	FLH181	SN7454N
9006	ZN7460E	9006	MC7460P	FJYIOI	DM8060N	FLYIOI	SN7460N
	ZN7472E	****	MC7472P	FJJHH	DM8540N	FLJIII	SN7472N
	ZN7473E		MC7473P	FJJ121	DM8501N	FLJ121	SN7473N
	ZN7474E		MC7474P	FJJ131	DM8510N	FLJ141	SN7474N
	ZN7475E		MC7475P	FJJ181	DM8550N	FLJ151	SN7475N
	ZN7476E		MC7476P	FJJ191	DM8500N	FLJ131	SN7476N
	21111102			FJJ141	DM8530N	FLJ161	SN7490N
				FJJ251	DM8532N		SN7492N
			MC7493P	FJJ211	DM8533N		SN7493N

#### RESISTOR TRANSISTOR LOGIC

RTI

General Characteristics

Gate speed 30ns, power 20mW Supply + Vcc 3.6V + 10% Frequency range 8MHz Noise immunity 0.3V

These RTL circuits are suitable for many simple switching operations where relatively medium speed operation is acceptable, for example, as an electronic relay.

#### **FAIRCHILD**

μL 910 Single inverter/driver T05 μL 914 Dual 2-input gate, positive NOR. negative NAND T05 μL 923 Single JK flip-flop with preset and clear T<sub>05</sub>

#### MOTOROL A

MC718P MC719P N MC722P MC723P		4-input OR/NOR gate Dual 2-input NOR gate Dual 3-input NOR gate Quad 2-input NOR gate Dual 3-input NOR gate Dual 4-input NOR gate JK flip-flop JK flip-flop Quad 2-input NOR gate JK flip-flop Quad 2-input NOR gate JK flip-flop Dual D-type flip-flop	MC780P MC785P MC787P MC788P MC789P MC790P MC791P MC792P MC793P MC793P	MC889P MC890P MC892P MC825P	Decade counter Quad 2-input expander I JK flip-flop, I inverter, 2 buffers Dual 3-input buffer, non-inverting Hex inverter Dual JK flip-flop Dual JK flip-flop Triple 3-input NOR gate Triple 3-input NOR gate Dual buffer Dual 4-input NOR gate
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Suffix G == TO5

plastic DIL

MC700 Series + 15°C to + 55°C

MC800 Series 0 to -- 75°C

#### DIODE TRANSISTOR LOGIC

DTL

General Characteristics

The following type characteristics apply to all manufacturers' DTL i.c.s using the coding given in the DTL tables.

Gate speed 25ns, power 5mW Supply + Vcc 5V ± 5% Frequency range 20MHz Input forward current - 10mA Input reverse current ImA Noise immunity 0.7V at 25 C Temperature range 0°C to 75°C Case outline DIL

DTL is suitable for most medium speed applications and is compatible with TTL. Positive logic usually applies to DTL circuit descriptions.

MC830P	Expandable dual 4-input NAND gate
MC831P	Clocked flip-flop
MC833P	Dual 4-input expander
MC836P	Hex inverter
MC846P	Quad 2-input NAND gate
MC848P	Clocked flip-flop
MC849P	Quad 2-input NAND gate
MC851P	Monostable multivibrator
MC856P	Dual JK flip-flop
MC862P	Dual 2-input NAND gate and inverter
MC886P	Dual 4-input expander

Temperature range - 55°C to - 125°C

#### Common Code Digits

Dual 4-input NAND gate (expandable) fan-out 8 Dual 4-input NAND driver gate (exp) fan-out 25

933 Dual 4-input expander

Hex inverter (expandable) Hex inverter fan-out 8

944 Dual 4-input NAND with open collector, lamp/ relay driver fan-out 27

RS flip-flop with preset and clear (master/slave) 945 fan-out 14

945 Quad 2-input NAND gate fan-out 8

948 RS flip-flop with preset and clear (master/slave) fan-out 13

Monostable multivibrator fan-out 10 951

962 Triple 3-input NAND gate fan-out 8

9093 Dual JK flip-flops with preset inputs

9094 Dual JK flip-flops with preset inputs

9097 Dual JK with preset inputs and common clear

9099 Dual JK with preset inputs and common clear

CASE OUTLINES (view from above case)

A -- T05 -- RTL μL900 B — T05 — RTL μL914 C — T05 — RTL μL923 D - Flat pack - 14 leads E - Dual-in-line - 14 leads F — Dual-in-line — 16 leads G — Dual-in-line — 24 leads











#### TRANSISTOR TRANSISTOR LOGIC

TTL

General Characteristics

The following type characteristics apply to all manufacturers' TTL i.c.s using the 74 series coding given in the TTL table.

Gate speed 10ns, nower 10mW

Fan-out 10

Supply + Fee 5V ± 5°a

Frequency range IOMHz

Input forward current - I-6mA (logic 0)

Input reverse current ImA

Noise immunity IV

Temperature range (74 series) 0°C to +70°C

Temperature range (54 series)  $-55^{\circ}$ C to  $+125^{\circ}$ C

Case outlines: mainly DIL (suffix N), flat pack, TO5 can. TTL is the most comprehensive range of logic i.c.s. It is used in high speed low power applications, where very short connecting lines (12in) can be used to avoid unwanted triggering from noise spikes. Positive logic applies to TTL circuit descriptions. TTL is compatible with DTL, ECL, MSI, using current sinking logic.

Suffix H indicates high speed (6ns) series, 22mW per gate, and fan-out of 8. Suffix L indicates low power (ImW) series, low speed (33ns), high fan-out 49.

Suffix N indicates a dual-in-line package although not shown in this list. Some suppliers omit the N in their lists so check before buying whether DIL or flat-pack is available: DIL is usually cheaper and easier to use. DIL package holders are recommended for easy insertion and removal of the i.c.

#### TTL GATES, BUFFERS, DRIVERS

7400 Quad 2-input positive NAND

7401 As 7490 but with open collector outputs

7402 Quad 2-input positive NOR

7403 As 7401 but pin connections as in 7400

7404 Hex inverters

7405 Hex inverters, open collector outputs

7406 Hex inverters, open collector buffer/drivers

7407 Hex buffers/drivers, open collector outputs

7408 Quad 2-input positive AND gates

7409 As 7498 but with open collector outputs

7410 Triple 3-input positive NAND gates

7413 Dual 4-input Schmitt triggered positive NAND gates

7420 Dual 4-input positive NAND gates

7430 Single 8-input positive NAND gate

7440 Dual 4-input positive NAND buffer

7450

Dual two-wide, 2-input AND-OR-INVERT (expandable)

7451 As 7450 but not expandable

7453 Single four-wide, 2-input AND-OR-INVERT (expandable)

7454 As 7453 but not expandable

7460 Dual 4-input expander for 7450 or 7453

7486 Quad 2-input exclusive-OR gates





#### TTL FLIP-FLOPS

7470 Single edge-triggered JK with preset and clear

7472 JK master/slave flip-flop with preset and clear

7473 Dual JK master/slave flip-flops with clear inputs

Dual D-type edge-triggered flip-flops with preset 74.74 and clear

74.75 Quad D-type 4-bit bistable latches with paired clocks (no clear)

Dual JK master/slave flip-flops with preset and 7476 clear

74107 Dual JK master/slave flip-flops as 7473, different connections

74118 Hex set/reset latch with common reset

74119 Hex set/reset latch with common and separate

74121 Gated or d.c. triggerable monostable multivibrator

#### MSI

#### TTL MEDIUM SCALE INTEGRATION

7441A BCD to decimal decoder/drivers

7442 BCD to decimal decoders

7445 BCD to decimal decoder drivers high power 30V

7445 BCD to seven segment display decoder/drivers open collector outputs 30V

7447 As 7446 but I5V output

7448 As 7445 but with passive pull-up outputs

7449 BCD to seven segment display decoder/drivers, open collector outputs, blanking input

7480 Gated full adder

7481 Sixteen-bit active-element read/write memory

7482 Two-bit binary full adder

7483 Four-bit binary full adder

7484 As 7481 but with gated write-amplifier inputs

7488 256-bit read-only memory

7490 Decade ripple counter with reset and preset to 9

7491A Eight-bit shift register (serial-in, serial-out)

7492 Divide-by-two and divide-by-six counter (+12) with reset

7493 Four-bit binary counter with reset (= 16)

7494 Four-bit shift register, dual entry parallel-in, serial-out

7495 Four-bit right-shift left-shift register

7496 Five-bit shift register, parallel-in, parallel-out

74100 Eight-bit bistable latches

74141 Improved version of 7441 to minimise switching transients

74145 Four-line to ten-line decoder, open collector outputs

16-bit multiplexer with strobe input 74150

74151 8-bit multiplexer with strobe input

74153 Dual 4-bit multiplexer with strobe inputs

74154 Four-line to sixteen-line decoder with strobe input

74155 Dual two-line to four-line decoder with strobe inputs

74156 As 74155 but with open collector outputs

Synchronous decade counter with parallel inputs 74160

74161 Synchronous binary counter with parallel inputs

74190 Synchronous up/down decade counter

Synchronous up/down binary counter 74191

74192 Synchronous up/down BCD counter

Synchronous up/down 4-bit binary counter

# APPLICATIONS OF LOGIC I.C.s

Fig. 1 shows a simple double pulse detector using a couple of DTL devices, especially handy for detecting index marks in position sensing equipment. If the input trigger pulses are derived from slowly

changing sources a couple of gates will be required ahead of the circuit shown to speed up the edges ready to fire the monostable (these are available anyway in the 946 range of DTL).

anyway in the 946 range of DTL).

Fig. 2 shows a scheme for a solid state stepping switch which could be useful for example for channel

selection in a radio or television receiver.

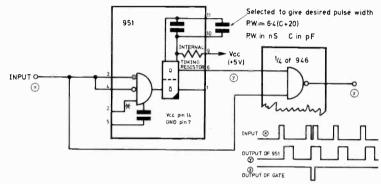
The operation is simple, the single push-button control is pushed an appropriate number of times until the required channel number is displayed on the readout. At this point the selected output of the 74145 is low, and can be used to drive a wide variety of external circuitry.

Many variations of this circuit are possible, and the whole thing can be made little larger than a cigarette packet. This example illustrates an interesting substitute for a wafer switch for some applications.

#### Fig. 1

#### **DOUBLE LINE PULSE DETECTOR**

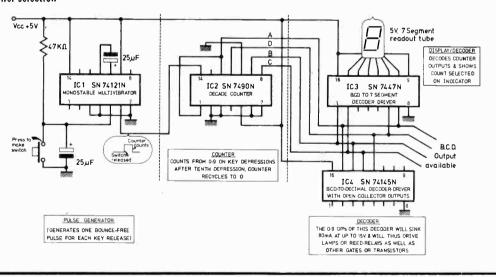
Gives an output pulse when a double pulse occurs. Monostable triggers on negative edge of input pulse. One application is to detect a double slot in rotating discs using a photo-electric cell



#### Fig. 2

#### SOLID STATE STEPPING SWITCH

This switch uses four TTL/MSI integrated circuits and provides seven segment digital readout for ten position channel selection



#### I.C. SUPPLIERS

The following advertisers in this magazine specialise in the supply of integrated circuits to readers:

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An example of a complex group of sunspots. This was an extremely active area on the sun which lasted for several days. The small spot below the main group is about the size of the earth. The magnitude of the group is apparent from this comparison.

# RADIO ASTRONOMY TECHNIQUES

By F. W. Hyde

PART 7

AVING now had some experience in the operation of the simple radio telescope, the "feel" of the subject will no doubt have been acquired. After the experience of sorting out what is interference, what is random noise, and what is the real signal to be acquired, the limitations of the simple unit will be apparent. However, although this article will begin the description of interferometers and details of their construction, it is recommended that the observer should continue with the programme of the first project for two reasons.

The first reason is that it takes some time to become facile at recognition of the many different forms of interference, some of which will be manmade and some random signals due to odd reflections from the upper atmosphere. The second reason is that the data acquired and entered in the log will be of considerable importance when compared with the measurements made with the interference.

ferometer.

Of course, some constructors will not have the space to set up an interferometer and so must in any case continue with the single aerial unit. Bearing this in mind, some suggestions that may help these people will be given in the last article of this series.

#### **IDENTIFICATION OF SPACE SIGNALS**

If the unit has been used during the night with the aerial pointed upwards, it is certain that there will be some signs of the radiation from the sources in Cygnus and Andromeda, and also the Crab Nebula. Also, the general rise in level of the recording at different times during each successive day and night will enable the positive identification of extra terrestrial radiation.

This was, in fact, the way in which the original pioneer Karl Guthe Jansky identified the existence of signals or radiation from space. To quote from his original paper, "In conclusion data have been presented which show the existence of electromagnetic waves in the Earth's atmosphere which apparently come from a direction that is fixed in

space. The data obtained give for the coordinates of this direction a right ascension of 18.00 hours and a declination of  $-10^{\circ}$ ."

Bearing in mind that the width of the aerial beam is quite wide in the vertical direction, tip the aerial so that the dipoles are directed at the Milky Way. Let the system run for a week every night or every three nights if this is more convenient. After the run compare the recordings by laying them side by side with the time marks aligned so that the same time is shown. Examine them for some significant event and check the difference in time that has elapsed. There should be a change of some 15 minutes in three days and some 30 minutes in a week. The exact times can be checked from the charts and the figures given are for the purpose of guidance only.

If it is possible obtain a copy of the *Ephemeris* issued by the Nautical Almanack Office. This may be available from the library or it can be obtained from the Stationery Office Bookshop in Holborn,

London.

Table 7.1 indicates some of the radio sources that should be detectable quite easily on the simple radio telescope. Since there are bound to be satellites recorded, it will be quite clear from the table which these are, because they can be compared with the tracings shown in Part 6 of this series.

#### SIDEREAL TIME

The table gives the usual astronomical coordinates of hour angle and declination. This involves certain corrections on the part of the observer for the hour angle or right ascension. This hour angle is in Sidereal Time which differs from ordinary astronomical time by an acceleration of 10 seconds in each hour, or approximately four minutes per day. The time is reckoned from the entry of the first point of ARIES which occurs between the twenty-first and twenty-third of March each year. This is the Vernal Equinox, 00.00 hours right ascension. All of this may appear a little complicated but it can be reduced to fairly simple terms for the purpose of the project.

In Table 7.2 a summary is given of the sidereal time for noon G.M.T. for each day of the year. This is approximate, but will not vary by more than about five minutes for each day and this error is insignificant with the system being used. In any case the source will be well within the beam of the aerial.

#### SIMPLE INTERFEROMETER

The simple two-aerial interferometer was briefly described in Part 2 of this series, and the modification required to convert the existing simple telescope is really one of addition. Firstly the existing aerial must be duplicated, together with the preamplifier if this is located at the aerial. The two aerials should be set up preferably on an east to west baseline. Within certain limits this is not an absolute requirement, but this will be dealt with later in this article. The distance between the aerial centres should be, if possible, at least five wavelengths. At the frequency that has been adopted this will mean a distance of 36 feet.

It is necessary to connect the aerials together. This may be done at the observing point, or if this distance is greater or about the same length as the distance between the aerials, then the following

alternative arrangement is possible.

Having set up the aerials, cut two lengths of cable which will leave sufficient slack for the aerials to be moved further apart and to facilitate stowage. These

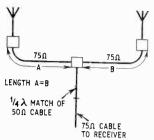


Fig. 7.1. Coupling two aerials together as a simple interferometer

two lengths of cable must be exact to within about one half inch, and of the same type of cable. These two points are very important. As both ends of the cables will terminate in coaxial plugs the equality of length includes the plugs. At the junction of the two aerial cables there will be a connection box as shown in Fig. 4.6 of Part 4. The outlet from this box will go to the converter in the same way as for the simple telescope, that is, with the matching section of 50 ohm cable followed by the 75 ohm cable. The diagram in Fig. 7.1 gives the layout.

cable. The diagram in Fig. 7.1 gives the layout.

The importance of the length of the cables has been stressed because the operation of the two aerial system depends on the combining of the signals which are alternately in phase and out of

phase with the other.

CONSTELLATION	I.A.U.	RIGHT	DECLINATION	10 Palmina and
	NO.	ASCENSION	DECEMBRICA	IDENTIFICATION
*Cassiopeia	00N6A	00h 22m ± 2m	64° 15′ ± 35′	(12) Supernova 1572 (no visible
Andromeda	00N4A	00h 40m 15s ± 30s	40° 50′ ± 20′	remnants) (11) M31 NGC224
Perseus	03N4A	03h 16m 37s ± 4s	41° 25′ ± 6′	(17) NGC127F C 111 11
Auriga	04N4A	05h 08m ± 4m	46° ± 1°	(17) NGC1275 Colliding galaxies? (13) Angular size of source is 1° 4 Identified with galactic nebu-
*Taurus	05N2A	05h 31m 35s ± 5s	22° 04′ ± 5′	losity
Orion	05S0A	05h 33m·0 ± 0m·2	-05° 27' ± 5'	(4) MI NGC1952 Crab Nebula
Gemini	06N2A	06h 13m 37s ± 4s	22° 38′ ± 5′	(18) Orion Nebula M42 NGC1976
Monoceros	08S0A	08h 08m ± 10m	22 38 = 5	(19) IC443
Lynx	08N4A	08h 09m ± 2m	-06° ± 30°	(13)
Puppis	08\$4A	08h 20m ± 4m	48° ± 1°	(8)
Lynx	09N4A	09h 16m ± 4m	-42° 30′ ± 1°	(13) Galactic nebulosity
Hydra	09SIA	09h 16m ± 2m	47° ± 1°	(8)
Ursa Major	09N6A	09h 51m 20s ± 2m	-12° ± 2°	(13)
Vela	10S4A	10h 10m ± 4m		(15) NCG3031
Ursa Major	ION5B	10h 30m ± 2m·5	$\frac{-42\frac{1}{2}^{\circ}\pm20'}{57^{\circ}\pm2^{\circ}}$	(2)
Crater	IISIA	11h 38m ± 8m	-15° ± 2°	(11)
Canes Venatici	12N4A	12h 15m ± 3m	47° ± 14°	(13)
Virgo	IZNIA	12h 28m 11s ± 37s	120 41 2	(11) NGC4258
Centaurus	13S4A	12h 22m 24s ± 1m	12° 41′ ± 10′	(I) M87 NGC4486
Canes Venatici	13N4A	13h 27m 30s ± 3m	-42° 37′ ± 8′	(I) NGC5128
Boötes	14N5A	14h 10m ± 2m	47° ± 1°	(11) NGC5195
Serpens Caput	ISNIA	15h 10m ± 4m	51° 30′ ± 1°	(11) NGC5457
Triangulum Australe	I6S6A	16h 10m ± 8m	11° ± 1½°	(2)
Hercules	16N4A	16h 36m ± 10m	-601° ± 5'	(2)
Hercules	16N0A	16h 45m ± 2m	41° ± 2°	(13)
agittarius	17S2A	17h 42m ± 1m	6° ± 1°.5	(2)
			-28°·5 ± 0° <u>i</u> 2	(21) May be associated with the galactic nucleus. The presence of neighbouring intense emission regions makes the measurements
Ophiuchus	18SOA	18h 16m ± 4m	—8° ± 2°	of flux density difficult
agittarius	18S1A	18h 17m·9 ± 0m·2	-16° 14' + 5'	
Cygnus	19N4A	19h 57m 44s ± 24s	40° 35′ ± 14′	(18) Omega Nebula M17 NGC6618 (3) Colliding galaxies
Cygnus	20N4A	20h 22m	40° · 1	(11) Cyg X extended source possibly associated with galactic
Cassiopeia	23N5A	23h 21m 36s ± 30s	58° 35′ + 10′	nebulosity (11) Galactic nebulosity

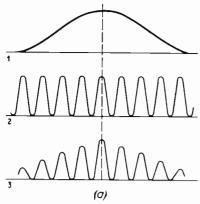


Fig. 7.2a. The effects of differing aerial parameters. (1) Single aerial pattern. (2) Two aerial pattern. (3) Interferometer pattern (point source)

When using an interferometer it is usual to have the aerials aligned and fixed in azimuth facing south in the northern hemisphere and north in the southern hemisphere. The altitude will depend on the section of sky to be observed, but each will normally be at the same angle of altitude. This will apply to each sweep of the sky. The sweeping will be done by the rotation of the earth so that in the course of 24 hours the whole 360 degrees of sky, as seen from the earth, will be scanned by the width of the aerial beam.

#### **RESOLUTION**

The resolution or ability to distinguish between sources will be dependent on two parameters. One is the distance between the aerials, for this governs the width of the fringes, and the width of a fringe

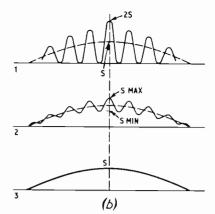


Fig. 7.2b. (1) Point source pattern. (2) Uniform extended source comparable but smaller than lobe spacing. (3) Uniform extended source equal to or greater than lobe spacing

determines the size of the smallest source that can be recognised. The diagrams in Fig. 7.2a and b illustrate the effects of these varying conditions.

It is essential that the beginner goes through this step of the simple interferometer in order to make practical observations and again have experience of the working conditions.

#### FRINGE WIDTH

Before moving on to the more complicated phase-switched interferometer there are one or two points which need to be covered. Firstly, the simple calculation to be made to determine the fringe width. The relationship between the fringes and the distance between the two aerials is in this form. The beam width between first null points (the fringe spacing) is  $1/D\lambda$  radians which equals  $57 \cdot 3^{3}/D\lambda$ .

TABLE 7.2

MONTH												
DAY	JAN.	FEB.	MAR.	APR.	MAY	JUNE	JULY	AUG.	SEPT.	OCT.	NOV.	DEC.
ı	18 40	20 43	22 33	00 35	02 33	04 36	06 34	08 36	10 38	12 37	14 39	16 37
2	44	46	37	39	37	40	38	40	42	41	43	41
3	48	50	41	43	41	43	42	44	46	44	47	45
4	52	54	45	47	45	47	46	48	50	48	51	49
5	56	58	49	51	49	51	50	52	54	52	55	53
6	19 00	21 02	53	55	53	55	54	56	58	56	59	57
7	04	06	57	59	57	59	58	09 00	11 02	13 00	15 02	17 01
8	08	10	23 00	01 03	03 01	05 03	07 01	04	06	04	06	05
9	12	14	04	07	05	07	05	08	10	08	10	09
10	. 16	18	08	11	09	- 11	09	12	14	12	14	13
11	20	22	12	15	13	15	13	16	18	16	18	17
12	24	26	16	18	17	19	17	19	22	20	22	20
13	28	30	20	22	21	23	21	23	26	24	26	24
14	32	34	24	26	25	27	25	27	30	28	30	28
15	35	38	28	30	29	31	29	31	34	32	34	32
16	39	42	32	34	33	35	33	35	37	36	38	36
17	43	46	36	38	36	39	37	39	41	40	42	40
18	47	50	40	42	40	43	41	43	45	44	46	44
19	51	53	44	46	44	47	45	<del>4</del> 7	49	48	50	48
20	55	57	48	50	48	51	49	51	53	52	54	52
21	59	22 01	52	54	52	54	53	55	57	55	58	56
22	20 03	05	56	58	56	58	57	59	12 01	59	15 02	18 00
23	07	09	24 00	02 02	04 00	06 02	08 01	10 03	05	14 03	06	04
24	- 11	13	04	06	04	06	05	07	09	07	10	08 .
25	15	17	08	10	08	10	09	11	13	11	13	12
26	19	21	11	14	!2	14	12	15	17	15	17	16
27	23	25	15	17	16	18	16	19	21	19	21	20
28	27	29	19	22	20	22	20	23	25	23	25	24
29	31		23	26	24	26	24	27	29	27	29	28
30	35		27	29	28	30	28	30	33	31	33	31
31	39		31		32		32	34		35		35

SIDEREAL TIME FOR NOON G.M.T. FOR 0° LONGITUDE

Using three examples of spacing, the first five wavelengths, the second 10 wavelengths, and a third 20 wavelengths, the approximate figures are 11.5 degrees, 5.73 degrees, and 2.86 degrees respectively. Since this is the point of the nulls the actual fringe width is half each value given at the half power points. The half power points were explained in Part 1 and illustrated in Fig. 1.4.

#### **CHART SPEED**

The second point to be noted is the appropriate speed of the chart on the pen recorder consistent

with obtaining a useful trace.

Taking the case of the least spacing 11.5 degrees, it will take 46 minutes for passing from the one null to the next, since it takes about four minutes of time for the earth to move 1.0 degree of arc. Therefore in this instance it will be wise to choose a slow speed for the recorder in order that the fringes may be clearly visible. A preferred speed would be 0.5 inches per hour, though 1.0 inch per hour would be permissible. For the widest spacing quoted a speed of 30 inches per hour would give best results. It will be clear from this that the greater spacing offers a better chance of detecting sources than the lesser spacing.

#### THE BASE LINE

Finally the setting up of the base line must be considered—for those who have the space to accommodate the interferometer. It is not sufficient to set up the south point using a compass, though this is a first step. The south point required is geographical south and this differs by several degrees. The exact amount of difference is obtainable for each year from ordnance survey maps. It was 8 degrees 40 minutes West in 1948, and decreasing by about 8 minutes annually.

Next month's article will be concerned with the phaseswitched interferometer.

## **NEWS BRIEFS**

#### TRAFFIC SPEED METER

COMPLETELY new road traffic speed measurement A system is being developed by GEC-Marconi Electronics, under contract to the Director of Telecommu-

nications, Home Office.

The new system uses an optical method of measurement and the complete system can be contained in a single unit which can be placed at the side of the road. The unit looks at right angles to the traffic flow and as soon as a vehicle passes, and comes within the field of view of the optical system, its speed will be measured almost instantaneously, and shown on a three-digit display, probably using a liquid crystal system.

The image of the vehicle is split into a succession of vertical strips, which are viewed by a photodiode. Any irregularities in the image, whether bright spots or shadow, will move across the slits of the "virtual" grat-ing and produce a fluctuation of the light falling on the diode. The frequency of this fluctuation is measured and, from it, solid-state micrologic circuitry calculates

the speed of the vehicle.

# EUROPE'S FIRST FULLY AUTOMATIC NAVIGATIONAL

N ALMOST every field nowadays man is rapidly being replaced by electronic equipment. This is the case at Portland Bill, Dorset, where a fully automatic navigational buoy has recently replaced the manned

Shambles lightship.

The 84 ton buoy, made by Hawker Siddeley Dynamics and named Lanby (Large Automatic Navigational Buoy) has a 40ft lattice mast topped with a main light beacon giving a luminous range of 16 miles. Also on board is a powerful fog signalling device, with an audible range of more than 3 miles.

The buoy is monitored every 30 minutes by a shore control station using a radio telemetry link. Should any failure occur, standby services operate automatically, and the nature of the fault is relayed to the control

station.

The buoy can be moored in depths from 30 to 300ft. and can withstand winds up to 100 mph, waves up to 40ft and tidal currents up to seven knots.

#### UNDERWATER TELEVISION EQUIPMENT

THE present system of searching for and locating I underwater wreckage, such as crashed aircraft and sunken ships by use of drag lines towed by surface craft seems soon to be superseded by a more efficient electronic method. This is evident following recent trials carried out by EMI Systems & Weapons Division, Middlesex, with their newly developed "low light" television equipment.

The television camera can quickly scan large underwater areas where visibility is poor and transmit clear pictures of the sea bed to a mother ship on the surface where they can be viewed and recorded. The television equipment will be housed in a midget submarine capable

of depths down to 3,000ft.

The "low light" television camera used in the trial is claimed to be the smallest in Britain, measuring 128  $\times$  128  $\times$  152mm. Its ability to obtain clear pictures in very low lighting conditions is provided by an E.M.l. Ebitron tube. This intensifier-vidicon is 300 times more sensitive than a normal 26mm vidicon tube.

#### COMPUTER NURSE

THE United Birmingham Hospitals have received approval from the Department of Health and Social Security to proceed with the purchase of a UNIVAC 418-111 computer as part of the Department's experimental programme in hospital computing. Delivery will be made early in 1972 and specially designed accommodation to house it is nearing completion at the Queen Elizabeth Medical Centre. The first real-time application is expected to go live early in 1973.

The initial configuration will provide for complete patient administration and laboratory reporting systems for the Queen Elizabeth Hospital. It is planned to introduce other experimental systems such as laboratory requesting, drug prescribing, nursing records, physical examination of the patients and other medical applica-

tions at later stages in the project.

#### NORTH MIDLANDS READERS Please note

A short course of six evening leccures entitled "Practical Electronics" will be held at North Staffordshire Polytechnic, College Road, Stoke-on-Trent, commencing Wednesday, January 12. The lecturers include well-known contributors to this Magazine.

Further details will be given next month.

#### SPECIAL PURCHASE!

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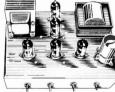
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QUALITY RECORD PLAYER AMPLIFIER MK II
A top-quality record player amplifier employing heavy
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Complete with output transformer matched for 3 ohm
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DE LUXE QUALITY PORTABLE R.P. CABINET MK II Uncut motor board size 141 12in., clearance 2 in. below, 54in. above. Will take above amplifier and any B.S.R. or GARRARD changer or Single Player (except A760 and SP25). Size 18×15×8in. PRICE 24.75. P. & P. 50p.

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A stylishly finished monaural amplifier with an output of 14 watts from 2 EL84s in push-pull. ELS4s in push-pull, Super reproduction of both music and speech, with negli-gible hum. Separate inputs for mike and gram allow records and announcements to follow each other



to foilow each other Fully shrouded section wound output transformer to match 3-15Ω speaker and 2 independent volume controls, and separate bass and treble controls are provided giving good lift and cut. Valve line-up 2 ELE4s, ECC33, EF66 and EZ80 rectifier. Simple instruction booklet 139 (Free with parts). All parts sold separately, ONLY 27-97, P. & P. 55p. Also available ready built and tested complete with std. input sockets, 29-97, P. & P. 55p.

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A.c. mains 200-240 volts. Using heavy duty fully isolated mains trans-former with full wave rectification

rectification giving adequate smoothing with negligible hum.
Valve line up:—2×
21. EZ80 as rectifier. Two dual potentiometers are provided for bass and treble control, giving bass and treble boots and cut. A dual volume control is used.
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LATEST B.S.R. C109/A21 4-SPEED AUTOCHANGER.

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TOMSTARIC COMPATIBLE STEREO CAETRIDGE
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LATEST ROBETTE T/O Stereo Compatible Cartridge for
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Designed for Hi-Fi reproduction of records. A.C. Mains operation. Ready built on opiated heavy gauge metal chaesis, size 75 in w. × 4 in. d. × 44 in. b. Incorporates ECC83, EL64, EZ80 valves. Heavy duty, double wound mains transformer and output transformer and output transformer matched for 3 obm wide range tone controls giving bass and treble lift and cut. Negative feedback line. Output 44 watts. Front panel can be detached and leads extended for remote mounting of controls. Complete with knobs, valves, etc., wired and tested for only \$4.75. P. & P. 35 p.

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MARVENSON'S SUPER MUNO AMPLIFIER
A super quality gram amplifer using a double wound fully
isolated mains transformer, rectifier and ECL82 triode
pentode valve as audio amplifer and power output stage.
Impedance 3 ohms. Output approx. 3-5 watts. Volume
and tone controls. Chassis size only 7in. wide - 3in. deep x
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#### Din 20 Kit

20 watt, high fidelity loudspeaker kit contains all parts necessary to complete the system, except timber and other material for the cabinet itself, with detailed,

illustrated instructions.
Specification: 20 Watts DIN,
4 ohms impedance, 8 ins
bass unit, dome HF
radiator, crossover
frequency 4,000 Hz.





#### Axent 100



Dome HF Radiator with integral crossover.
Capable of high frequency sound reproduction with negligible distortion in systems rated up to 30 Watts DIN, this 'state of the art' drive unit has an integral crossover which cuts frequencies below 3kHz at a rate of 12dB/Octave.

#### Audiom 100

12 inch high fidelity bass loudspeaker.
For use as a bass unit in two-way systems, the

two-way systems, the sensitivity and high frequency roll-off of the Audiom 100 has been tailored to match the Axent 100.



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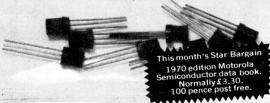
All fully coded, all from well-known manufacturers and now available, while stocks last, at better than bulk-buyer's prices! Cash with order only.

THIS MONTH: **NPN Silicon Transistor** 2N2926 (Red) Hfe 55-110 8 for 50p C6U 1.6 amp general purpose 25V SCR in TO5 case 3 for 50p D16P4 (Equals 2N5306) Darlington transistor Hfe min = 7000 3 for 50p Silicon NPN ultra high gain 2N3390 transistor Hfe 400-800 3 for 50p 2N3391A Silicon high gain low noise transistor (better than BC109) 3 for 50p

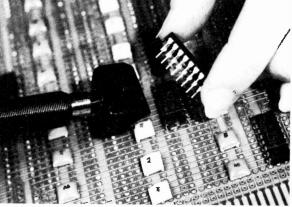
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# JUST A SECOND HOME CONSTRUCTOR



Yes, about a second, to remove dual-in-line I.C. and T.O.5 type transistor from those ex-computer printed-circuit boards, experimental circuits, etc.

DILKIT comprises a high performance soldering iron, long-life bits specially shaped to unsolder all leads of a component in a single operation, and a length of solder-absorbing wick.

Automatic component extractors also available.

TO: ORIENTATION LTD., MAYFIELD, COVERACK, CORNWALL.
PLEASE SEND

	DILKIT £7:40		
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	T.O.5 EXTRACTOR £2-25		
CHEQUE/P.O.		VALUE	

#### SURPRISE SURPRISE!!

The decision by RCA to pull out of the computer mainframe business sent a chill down the spine of the United States' electronics industry—and of Europe too.

Everyone was taken by surprise. It happened almost exactly a year after RCA chairman Robert W. Sarnoff had announced firm plans for a frontal attack on the market held so lucratively by IBM. As recently as July, Sarnoff was still denying that RCA had any intention of pulling out.

The RCA debacle follows a similar withdrawal by U.S. General Electric a year earlier. GE's move, however, was well regulated and involved lengthy negotiations before the sale of its computer

interests to Honeywell.

Industry experts on both sides of the Atlantic are asking themselves: if giants like U.S. General Electric and RCA can't stand the pace, who can? Except, of course, the

almighty IBM.

Unstoppable IBM has 70 per cent the U.S. domestic market and not far short of that figure in the rest of the world. RCA's share was under 4 per cent with a growth target of 10 per cent of the market by 1975. The crunch came when it was realised that another  $$\Sigma$200$  million of capital, on top of the  $$\Sigma$200$  million already spent, would be required before breaking into profit.

Now the RCA market share will be distributed among the survivors with the bulk going to IBM, still further tightening IBM's strangle-

hold.

#### THE AFTERMATH

Britain's ICL is involved only through System 4, a development of RCA's Spectra 70 Series. ICL say they will remain unaffected. But Siemens in Germany will be affected because they rely strongly on close technical links with RCA.

on close technical links with RCA.
RCA customers have been told that all existing orders and service contracts will be honoured. But the demise of two major mainframe manufacturers in a period of 16 months will influence new buyers toward the most stable source which means even more business for IBM.

Component suppliers to RCA such as Motorola, Signetics and Advanced Memory Systems will all suffer from the closure. Some smaller suppliers stand to lose 10 per cent of their total business.

At the time of writing there was no firm news that RCA had been able to sell its computer interests either in whole or in part. In the present depressed state of the computer market few firms are interested in acquiring extra plant and manufacturing capacity.

One possibility emerging from the debacle is that Britain's ICL and Germany's Siemens could now get together to form a new and potentially extremely powerful computer force in Europe.

#### DATA SATELLITE

The computer industry may be suffering at the moment but the demand for data processing and data transmission will continue to increase at a growing rate. In 1970 there were 14,000 data terminals in Britain. By 1980 there might be, and probably will be, 300,000. This figure will be matched by other major European nations and so far no international agreements on data network standards have been made.

An interesting new development is a study contract awarded to Marconi Communications Systems Ltd. by the European Space Research Organisation (ESRO). The idea is that data links should be through a geo-stationary satellite used in conjunction with large numbers of small and inexpensive

earth terminals.

The earth terminals would have dishes no larger than 9ft in diameter sited within 100yds of the data terminal. Use of such a system, dedicated entirely to data transmissions, would obviate many of the difficulties of international connections through existing public telephone and data networks where data rates are comparatively low due to line variations.

The preliminary study, involving analysis of present and future traffic, prospective users and costs of the system, should be com-

pleted by April 1972

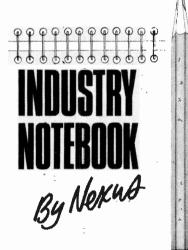
#### WHERE THE GIANTS MAY GO

Although the great electronic exhibitions aimed at industry such as the IEA and ILECS at Olympia are by no means on the way out, increasing costs have been forcing companies to look more closely at the less expensive small shows which tend to be more cost-effective.

One of the brightest operators in the field is John McNeill, managing director of Electromation Exhibitions Ltd. I have just seen the company's forward programme with electronics shows starting at Birmingham in November and moving round the country to Harrogate, Southend, Stevenage, Portsmouth, Croydon, Glasgow. Dublin, Bristol, Manchester and back to Birmingham in November 1972. This is an impressive programme which includes, apart from exhibitor stands at each venue, a technical convention. Admittance is reserved for professional engineers and you need an invitation before you can get admission.

Another bright operator in small exhibitions is Evan Steadman who has already sold out for his Seminex due to be held next April. This one is for semiconductor manufacturers. Yet another is Gordon Johns who organised the successful Compec '71 computer peripheral exhibition in London last September.

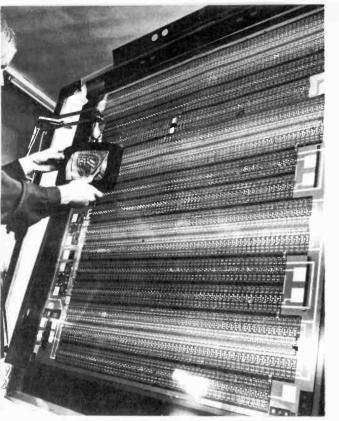
It is too early to predict the effect of this type of exhibition on the IEA at Olympia next year but one thing seems certain. With Mullard, TI, Fairchild, RCA, Emihus, Ferranti, to name but a few who have already booked for Seminex and its accompanying technical symposium, it seems clear that the bulk of the semiconductor companies have already opted out of the big show.



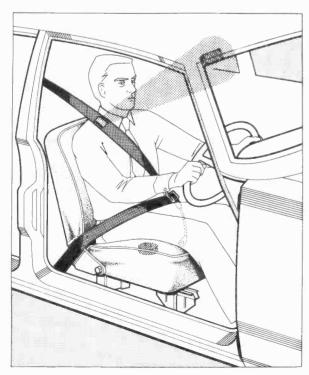
#### BRIGHTER OUTLOOK— BUT SLOW

Sir John Clark, chairman of Plessey, thinks the worst may be over but it will be a long haul back to the halcyon days of prerecession. Ernie Harrison, Racal chairman, believes that it could be another 18 months before business takes off in any big way. Dr F. E. Jones, managing director of Mullard was not exactly cheerful when we last met despite the boom in colour TV sales.

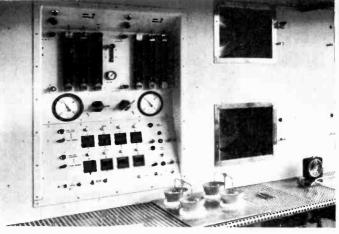
On the whole, however, I have found most industry leaders optimistic. Smaller companies with their greater level of flexibility have in many cases proportionately weathered the storm better than the giants. But 1972 could see further "re-structuring" of the industry through takeover or merger. On one thing they all agree. Even if the bottom has now passed, don't expect a dramatic revival. Recovery will be slow.



Physical inspection of rubyliths (artwork) of a 1,024-bit CDI shift register



Layout of the car safety belt system as installed in a car. Note the narrow ultrasonic beam directed at the receiver mounted in the windscreen pillar



Part of the pre-production line for CDI

# New LSI Combines Logic and Linear Functions on one chip

How does the thought of a high performance computer on a single slice or even a chip strike you—impossible? Well it is not as far fetched as it may seem. While we are just getting accustomed to MOS integrated circuit technology, along comes another—the CDI process of producing LSI (large scale integration).

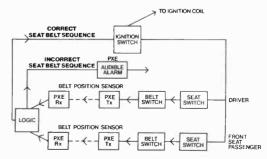
CDI (collector diffusion isolation) is a simple bipolar process originally developed in the Bell Laboratories in the U.S.A. and has now been further developed by Ferranti at Manchester for large scale

# ELECTRO

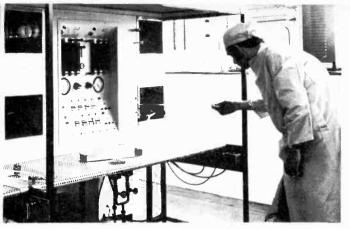
#### Ultrasonic Car Safety Belt System

ONE of the results of a new car safety belt system, developed by Mullard and the Ford Motor Co., means that even the car thief will have to strap himself in if he wishes to make off with his "booty". The belt must not only be fastened but also be correctly positioned across the wearer.

As the block diagram shows, before the ignition will function the driver must occupy his seat, activating a pressure switch fitted beneath it, buckle the belt across his lap and close the belt switch. Provided the belt is worn correctly a 40kHz signal emitted from an ultrasonic transmitter, mounted in the belt, is received by a detector built in the windscreen pillar, which in turn completes the ignition circuit. If the front passenger seat is occupied then a 50kHz signal must be received from the passenger's



Block diagram of the car safety belt system



bipolar integrated circuits at Ferranti

production at an economic cost. This is said to be the first time that LSI has become a practical proposition, combining the high complexity of MOS with the performance advantages of bipolar technology.

The breakthrough is in the combination of both high speed switching logic and linear capability for a common supply of 5V on the same monolithic chip. The system is completely compatible with existing TTL is s

existing TTL i.e.s.

Only five masking operations are required compared with nine steps involved in current bipolar processing. It is expected that this new bipolar LSI technology will be used where digital and analogue control, computers and telecommunications already require complex and sophisticated circuitry.



safety belt before the ignition switch will be functional.

If the correct sequence has not been carried out a logic circuit will trigger an audible and visible alarm mounted on the dashboard.

The system can be arranged so that if the belt is unfastened while the car is moving the ignition is not immediately affected. Instead, the alarm is sounded and if, at the end of a specified time, the belt still remains unfastened, the ignition will then be cut out.

For very short operations, such as parking or garaging the car, the logic arrangement can be adjusted to allow for car movement in first or reverse gear for a specified time without the driver being belted.

As a method of enforcing the wearing of safety belts by driver and passengers, thus cutting down the death and injury toll on the roads, this contribution to road safety should be closely looked at by the Police and the Ministry of Transport.

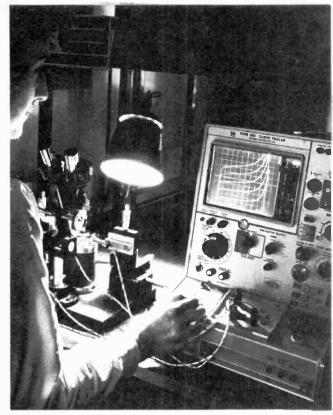






The ultrasonic transmitter (top) and receiver circuit boards

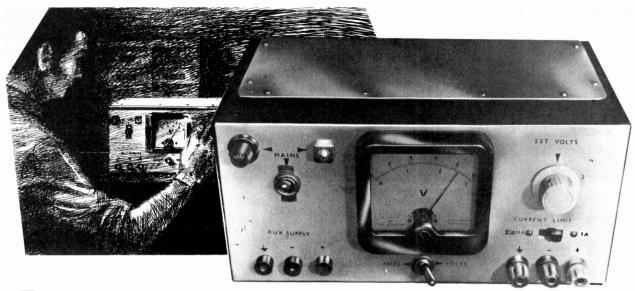
Practical Electronics December 1971



An operator carrying out a process control check using a curve tracer which shows the collector-emitter characteristics of a CDI transistor in a digital array

The ultrasonic transmitter mounted on a car safety belt





# Labpack BY H.T. KITCHEN Power Supply

THE stabilised power supply to be described was designed and built to provide a power source for transistorised equipment undergoing development. As such, it had to fulfil certain basic requirements.

A long and varied experience of development and service work had shown the need for a power supply capable of being varied from near zero to a maximum of some 25V, with a maximum current capacity, within this voltage range, of 1A.

A further requirement was that the output voltage had to be regulated. The output current also had to be controlled, so that even if the load terminals were made short circuit, the maximum current rating could not be exceeded.

The final requirement was that the maximum current output should be capable of being adjusted

Main D.C. Current Limiter Output Mains Supply TR5.TR6,TR7 TR3.TR4 Reference Differential Supply Amplifier 05,06 TR1,TR2 Auxiliary +ve D.C. Supply -Output D7, D8

Fig. 1. Block diagram of power supply

to values less than the 1A maximum. In the prototype, two switched ranges of 1A and 100mA are provided.

A block diagram of the power supply is shown in Fig. 1 which should be studied in conjunction with the circuit diagram of Fig. 2.

#### **SERIES REGULATOR**

The regulator element of this design is of the series type and is made up of the transistors TR5, TR6 and TR7.

This is a compound configuration where TR5 controls the base to emitter voltage of the output pair TR6 and TR7. The base feed resistor of TR5 also functions as the collector load of TR2. This transistor, together with TR1, constitute a differential amplifier.

#### DIFFERENTIAL AMPLIFIER

The differential amplifier provides an output voltage which is proportional to the difference of the input voltages applied to the transistor bases. At the base of TR1 is connected a fixed Zener voltage which is derived from a double diode rectifier and filter circuit.

This stable reference at one base means that any changes in the output voltage—either up or down—will be fed back by way of VR2 to provide a collector current change through R5. If the current flow is low the volt drop at this resistor is low, which corresponds to a movement of the base potential of TR5 towards the collector potential. Increased emitter current therefore flows, and as this current is into the bases of TR6 and TR7, their emitter currents also increase.

If the output load is a resistive one, this will result in an increase of output voltage. A converse

feedback action will decrease the output.

The overall gain of TR2, TR5 and the output pair is very high, and very small changes in output voltage can be sensed and automatically corrected. Rapid changes in output voltage, corresponding to a residual 100Hz ripple, or the signal frequency of a connected high power amplifier, can also be sensed and corrected. In this instance, feedback is via C5, again to the base of TR2 which treats it also as an error signal.

#### **CURRENT LIMITING**

The current limiter circuit comprises TR3 and TR4, in conjunction with VR3, R8 and VR4 and R10.

The transistor TR3 is connected so as to function as a cheap form of Zener diode which fixes the emitter potential of TR4. Under no load or small load conditions TR4 is reverse biased and does not pass any collector current.

Dependent on the switch position of S2 and limit setting of VR3 or VR4, load current limiting occurs when the voltage drop across these resistance combinations is sufficient to drive TR4 out of its cut-off state.

With TR4 conducting the volt drop across its shared load resistor R5 will increase. The polarity of this voltage is such as to reduce the  $V_{be}$  of TR5 and so reduces the emitter current passed by TR6 and TR7.

By controlling the  $V_{\text{be}}$  of TR5 it is possible to set a limit to the maximum current that can be drawn from TR6 and TR7, even if the output is short circuited.

#### CHOICE OF LIMITING RESISTANCE

Since any volt drop across R8 and R10 is lost to the output voltage, and since a certain minimum voltage at the base of TR4 is necessary for reliable current limiting, the absolute minimum resistance

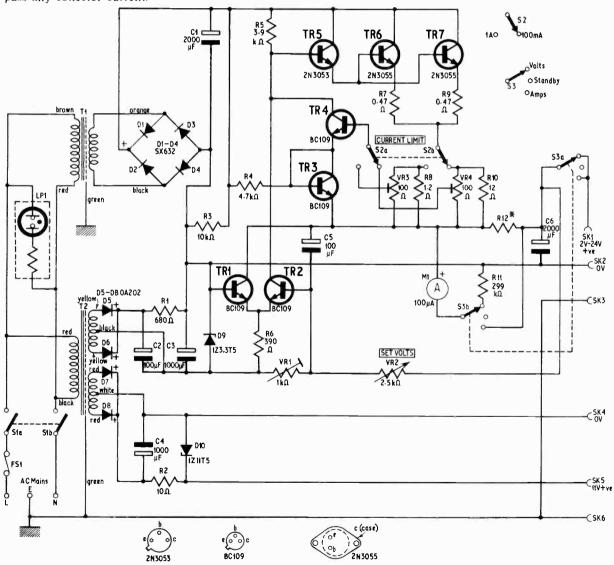


Fig. 2. Circuit diagram of power supply

### SPECIFICATION . . .



### R12 SK2 **М**ЕНТОНОО ОТ ТОТО #FOLTTTTTTTTTTTTTTTTTTTTT VR2 Rectifier 52b(A) S2b(mA) S2b (wiper) SZa(wiper) S2a (mA) 52a(A)

#### MAIN SUPPLY

Output Voltage Output Current **Current Limiting** Output Resistance

A.C. Ripple

A.C. Ripple
Output Variation

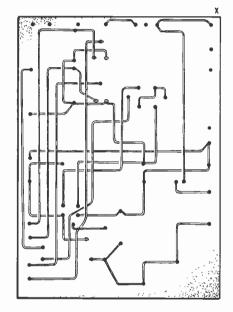
2V to 24V IA max IA and I00mA 50 milliohms at IA With no load is  $250\mu V$ 

At 20V, IA is 5mV r.m.s. 0.1% for 10% mains variation

#### **AUXILIARY SUPPLY**

**Output Current** Output Voltage

30mA max IIV  $\pm$  5% at 30mA



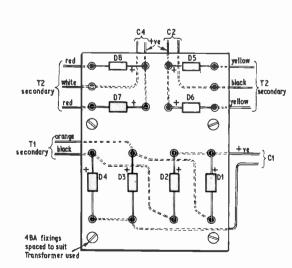
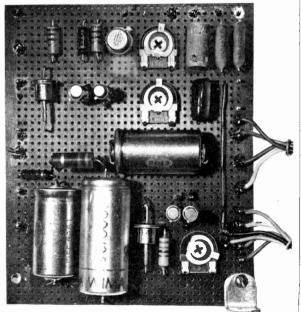


Fig. 4. Component layout and wiring of rectifler panel



values must be selected. The exact values used may require some adjustment, dependent on the gains of the transistors used in individual versions.

The action of the limiting circuit is not an abrupt one. The limiting process on the 1A range begins at about 0.8A and gradually increases. This will affect the output resistance above and below about 1A.

#### **AUXILIARY SUPPLY**

The transformer T2 has two windings, one of which provides the reference voltage for the differential amplifier, the other provides a stabilised 11V

#### COMPONENTS...

#### Resistors 680Ω R7 Ri 0.47Ω IW wirewound $\Omega$ 01 R8 1.2Ω 2.5W wirewound 0·47Ω IW wirewound $l0k\Omega$ R9 R3 **R4** 4·7kΩ RIO 12Ω 2.5W wirewound $3.9k\Omega$ RII 299k $\Omega$ 1% metal film (see text) R5 $390\Omega$ R12\* See text R6 All 5%, 4 watt carbon except where shown

Capacitors

ĊΙ	$2,000\mu$ F elect. $50V$	C4	1,000µF elect.	50V
C2	100μF elect. 50V	C5	I00μF elect.	50V
C3	1 000 Felect 50V		2 000 F elect	

Transistors
TRI-TR4 BC109 (4 off) TR5 2N3053 TR6-TR7 2N3055 (2 off)

#### **Diodes**

DI-D4 SX632 (4 off) D5-8 OA202 (4 off)
D9 IZ3-3T5 IW 3-3V Zener DIO IZIITS IW IIV Zener

#### **Potentiometers**

VRI IkΩ miniature horizontal preset VR2  $2.5k\Omega$  multi-turn potentiometer (see text) (R.S. Components) VR3-VR4  $100\Omega$  miniature horizontal presets (2 off)

#### **Switches**

SI Double pole mains on/off switch

Double pole, double throw slide switch

S3 Double pole, double throw "centre-off" toggle switch

#### Meter

MI 100µA Type MR26 (R.S. Components)

#### **Transformers**

TI Mains transformer: Primary 0-240V Secondary 26V at 1.6A

Type X6705 (Belclere Ltd) T2 Mains transformer:

Primary 0-240V Secondaries 26-0-26V, 23mA and II-0-IIV, 33mA

Type X676X (Belciere Ltd)

#### Miscellaneous

LPI Mains neon indicator, FSI 500mA fuse with holder,  $3\frac{3}{4}$ in  $\times$   $4\frac{3}{4}$ in pegboard 0·lin pitch,  $2\frac{1}{2}$ in  $\times$   $3\frac{1}{2}$ in pegboard, 0·l5in pitch, 20 s.w.g. aluminium sheet cut as required

Heat sink 4.875in × 1.05in extruded aluminium 4in long with eight pairs of fins, SKI-SK6 insulated

sockets (6 off)

at 33mA and can be used for powering any permanently attached piece of equipment, provided it does not draw current in excess of the rated figure.

#### METER CIRCUIT

For monitoring purposes a fairly large  $100\mu A$  meter is used. The one on the prototype was arbitrarily scaled 0-3 and 0-10, which is convenient. since with the multiplier R11 and shunt R12 no scale marking is necessary for the ranges 0-30V and 0-1A

A two pole, two way, "centre-off" switch (S3) provides the measurement requirement. The centre position provides a standby facility which can be used when changing connections or reversing supply polarities.

The multiplier resistor R11, should ideally have a resistance of 299 kilohms and can be ordered with the meter. It is possible to use a 300 kilohm, 1 per cent type as the loss of accuracy is not severe.

The construction of the meter shunt will be dealt with later.

#### **OUTPUT TRANSISTORS**

The more knowledgeable constructor will have noticed the use of two 2N3055 transistors for the series regulator TR6 and TR7, when one would suffice.

For maximum dissipation to occur in these transistors requires the output to be made short circuit and is approximately 45W. Clearly, a good heatsink and ample ventilation are required, for even under normal load conditions the dissipation can be as much as 30W.

Since the unit was intended to be used for long periods at fairly high dissipations, it was decided to use two output transistors for the reliability offered.

#### CONSTRUCTION

The disposition of the components, whilst permitting the use of a relatively small cabinet, does result in a high packaging density.

Most of the semiconductors, capacitors and resistor are mounted on a  $3\frac{1}{2}$ in  $\times$   $4\frac{1}{2}$ in plain piece of 0-lin pitch pegboard as in Fig. 3. Veroboard can be used if desired.

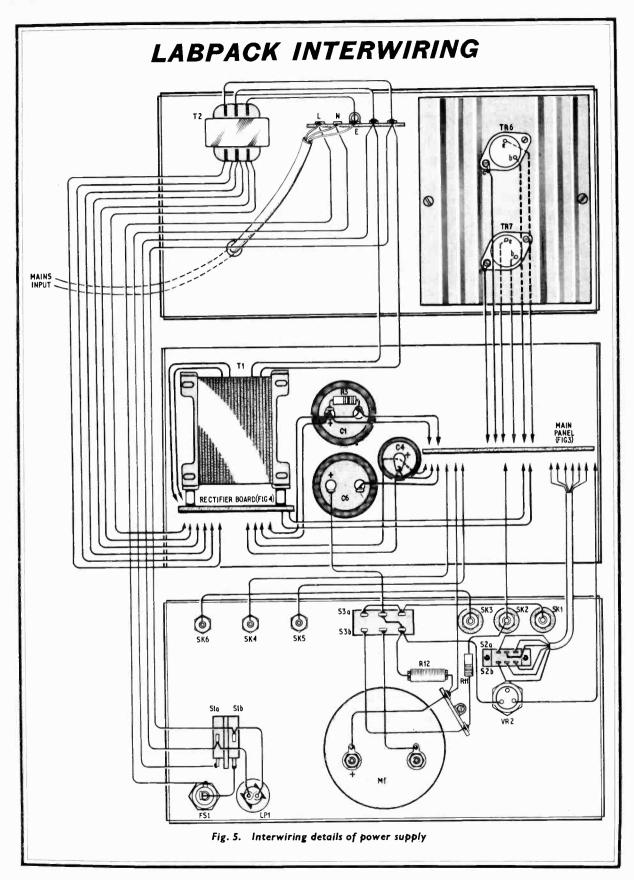
The rectifier panel of Fig. 4 is a 2½in × 3½in plain pegboard of 0.15in pitch. This should be drilled for mounting to the frame of T1 (see photograph). When the rectifiers are mounted the panel should be attached to the transformer using 4B.A. nuts and bolts and in spacers.

#### INTERWIRING

Interwiring details for the complete power supply are given in Fig. 5. The arrowed connections to the boards can be determined by referring to Figs. 3 and 4.

The front panel carries all the controls, the meter, fuse, indicator neon and output terminals (Fig. 5). On the prototype, the output voltage control, VR2 was a ten turn potentiometer, though somewhat expensive it is to be recommended since it enables precise setting to be achieved. Of course, an ordinary wirewound poter tiometer can be substituted.

Transformer T2 is one of the clamp type of construction and is secured to the rear wall of the case. So too is a five way tag strip, to which the mains input is connected, and the heat sink for TR6 and TR7.



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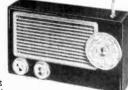
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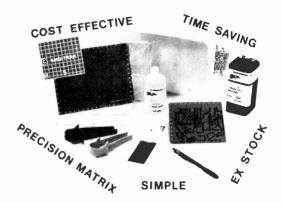
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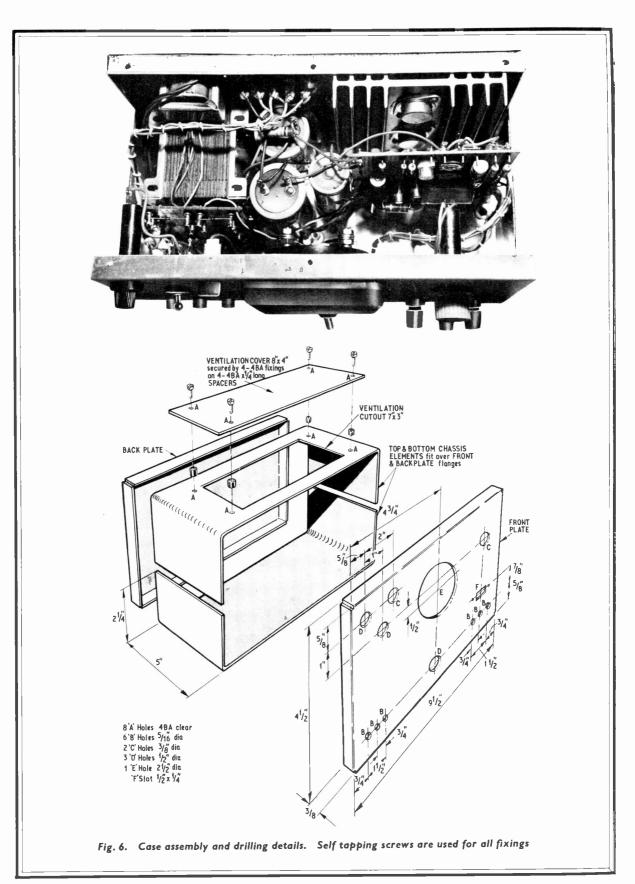
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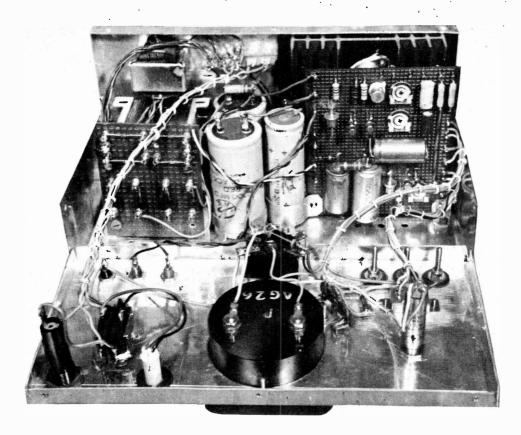
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Transformer T1 and capacitors C1, C4 and C6 are mounted on the cabinet base. The main component board is attached vertically at its two lower corners, one corner by means of a small right angled bracket, the other by the clamp of C4.

# MAKING THE CASE

The cabinet used for the prototype was made from 20 s.w.g. aluminium, bent and cut as in Fig. 6. In assembling, the various panels are all secured together by small self tapping screws.

Ventilation must be adequate and is effected by means of a rectangular hole  $3in \times 7in$  cut out of the top panel. This is covered by an aluminium plate  $4in \times 8in$  separated from the top by  $\frac{1}{4}in$  spacers.

To maintain air flow through this port  $\frac{1}{8}$  in holes should be drilled below the heat sink and around T1.

# **TESTING**

When all of the components have been assembled and connected, a careful check for wiring errors should be made. If everything appears satisfactory set VR1 and VR2 for maximum resistance, and VR3 and VR4 to mid-range. S3 should be set for a voltage reading. A load is not required.

With the unit connected to the mains and switched on, the voltmeter should indicate. VR2 should be rotated anti-clockwise, when the voltage should fall, then rise again as VR2 is returned to its original fully clockwise position.

The preset VR1 should now be set for a maximum output voltage of 24V.

# **METER SHUNT**

The meter shunt, R12, is made up of resistance wire with a measured value of 0.125 ohms. Since the formula for calculating this includes the meter resistance the meter specified, or one with a resistance of 1,250 ohms, must be used.

To actually construct the shunt, a length of resistance wire, somewhat longer than is required, is connected in the position of R12. With S3 set to "Volts" adjust the output of the unit for 10V. Switch off and connect a 10 ohm, 10W resistor, in series with a multimeter switched to 1A d.c., across the output terminals.

If S3 is switched to "Amps" and the unit switched on, R12 can be adjusted so that 1A flows in the multimeter when meter M1 registers full scale deflection.

A high value resistor can be used as a mounting for the resistance wire, it being simply wrapped round the resistor body with the wire ends soldered to the resistor leads.

# CURRENT LIMIT SETTING

Having set the output voltage, attention can now be turned to the current limiting potentiometers VR3 and VR4.

With S2 switched to the 1A range a multimeter switched to the 1A d.c. range can be connected directly across the output terminals and VR3 adjusted for a reading of 1A.

The time taken for this adjustment should be a minimum as the dissipation in the output transistors is high. Adjustment for the 100mA range is identical with suitable multimeter range switching. Here, of course, VR4 is adjusted.

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AC115 AC125	23p AF117	17p BC142	45p BCY33	17p BF274	30p MAT100	15p ST140	12n 2N1132	22p 2N2905A	30p 2N3	708 8p
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ACI27	17p AF125	20p BC147	17p BCY71	30p BF316	75p MATI21	17p UT46	27p 2N1304	20p 2N2907	25p 2N3	
ACI28 ACI4IK	17p AF126 17p AF127	20p BC148 20p BC149	12p BCY72 17p BCZ11	15p BFW10 20p BFX29	55p MPF102	43p V405A	25p 2N1305	20p 2N2907A	30p 2N3	BI9 40p
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AC151 AC154	15p AF178 15p AF179	50p BC151 50p BC152	10p BD123 17p BD124	85p BFX85	27p OC20	50p 2G302	19p 2N1308	27p 2N2925	13p 2N3	904 27p
ACI55	17p AFIBO	50p BC153	27p BD131	75p BFX86 80p BFX87	22p OC22 25p OC23	30p 2G303 33p 2G304	19p 2N1309 20p 2N1613	27p 2N2926 17p (G)	12p 2N3	
AC156 AC157	17p AF191	50p BC154	30p BD132	80p BFX8B	22p OC24	45p 2G306	35p 2N1711	20p 2N2926(	r) Hp 2N4	058 <b>15</b> p
AC165	17p AF186 17p AF239	45p BC157 37p BC158	20p BDY20 17p BF115	21 BFY50 22p BFY51	20p OC25 20p OC26	25p 2G308 25p 2G309	35p 2N1889 35p 2N1890	35p 2N2926 45p (O)	10p 2N40	
AC166	17p AFZII	37p BC159	20p BF117	45p BFY52	20p OC28	40p 2G339	17p 2N 1893	37p 2N3010	80p 2N40	
AC167 AC168	20p AFZ 12 20p AL 102	45p BC167 85p BC168	13p BF118 13p BF119	60p BFY53 70p BSX19	17p OC29 15p OC35	40p 2G339A 33p 2G344	15p 2N2160 15p 2N2147	60p 2N3011 75p 2N3053	20p 2N40	
AC169	14p AL103	85p BC169	13p BF152	35p BSX20	15p OC36	40p 2G345	15p 2N2148	60p 2N3054	20p 2N5 50p 2N5	
AC176 AC177	23p ASY26 20p ASY27	25p BC170 30p BC171	12p BF153 13p BF154	35p BSY25	15p OC41	20p 2G371	13p 2N2192	30p 2N3055	63p 2S03	4 75p
AC187	30p ASY28	25p BC172	13p BF157	35p BSY26 45p BSY27	15p OC42 15p OC44	22p 2G371B 15p 2G374	10p 2N2193 17p 2N2194	30p 2N3391 27p 2N3391A	17p 2S30 20p 2S30	
ACIBB ACYI7	30p ASY29 25p ASY50	25p BC173	13p BF158	25p BSY28	15p OC45	12p 2G377	27p 2N2217	20p 2N3392	17p 2S30	2 45p
ACYIB	20p ASY51	25p BC174 25p BC175	13p BF159 22p BF160	30p BSY29 30p BSY38	15p OC70 15p OC71	15p 2G378 9p 2G382	15p 2N2218 15p 2N2219	25p 2N3393 27p 2N3394	15p 2S30 15p 2S30	
ACY19	22p ASY52	25p BC177	17p BF162	30p BSY39	15p OC72	12p 2G401	30p 2N2220	22p 2N3395	20p 2530	
ACY20 ACY21	20p ASY54 20p ASY55	25p BC178 25p BC179	17p BF163	35p BSY40 35p BSY41	30p OC74 35p OC75	12p 2G414 15p 2G417	30p 2N2221 25p 2N2222	22p 2N3402 27p 2N3403	22p 2S30 22p 2S30	
ACY22	19p ASY56	25p BC180	20p BF165	35p B5Y95	12p OC76	15p 2N388	30p 2N2368	17p 2N3404	22p 2S30 32p 2S32	
ACY27 ACY28	18p A5Y57 19p A5Y58	25p BC181 25p BC182	22p BF167 10p BF173	22p BSY95A 22p BU105	12p OC77 63-90 OC81	25p 2N388A	50p 2N2369	15p 2N3405	45p 2S32	2 50p
ACY29	30p ASY58	25p BC182L	10p BF176	35p CITIE	60p OC81D	15p 2N404 15p 2N404A	22p 2N2369A 30p 2N2411	15p 2N3414 50p 2N3415	20p 2532 20p 2532	
ACY30	25p ASZ21 25p BC107	40p BC183 10p BC183L	10p BF177	35p C400	30p OC82	15p 2N524	55p 2N2412	50p 2N3417	37p 2S32	4 £1-20
ACY34	18p BC108	10p BC184	10p BF178 13p BF179	45p C407 50p C424	25p OC82D 17p OC83	15p 2N527 20p 2N696	60p 2N2646 12p 2N2711	55p 2N3525 22p 2N3702	74p 2S32 12p 2S32	
ACY35 ACY36	18p BC109	IIp BCI84L	13p BF180	30p C425	40p OCB4	20p 2N697	15p 2N2712	22p 2N3703	12p 2532	
ACY40	30p BC113 15p BC114	25p BC186 30p BC187	27p BF181 27p BF182	30p C426 30p C428	30p OC139 20p OC140	15p 2N698 17p 2N699	24p DIC	DES & R	ECTIFIE	DC
ACY41	18p BC115	30p BC207	IIp BFI83	30p C441	27p OC170	15p 2N706	7p		EUIIFIE	
ACY44 ADI40	35p BC116 40p BC117	35p BC209 35p BC209	IIp BFI84	25p C442 30p C444	35p OC171 37p OC200	15p 2N706A 25p 2N708	8p AA119 12p AA120	8p BYZII 8p BYZI2	32p OA8	
AD142	40p BC118	25p BC212L	Hp BFI88	30p C450	17p OC201	27p IN709	45p BAII6	22p BYZI3	30p OA8 25p OA9	
AD149 AD161	43p BC119 35p BC125	45p BC213L 35p BC213L	IIp BF194	23p C720 24p C722	12p OC202 25p OC203	27p 2N711 25p 2N717	40p BA126 42p BY100	22p BYZ16 15p BYZ17	35p OA9	l 7p
AD162	35p BC126	35p BC214L	12p BF196	30p C740	25p OC204	25p 2N718	24p BY101	12p BYZIB	35p OA9 30p OA2	
AD161/ 162(MP)	63p BC132	25p BC225 30p BC226	25p BF197 35p BF200	35p C742 45p C744	17p OC205 17p OC309	35p 2N718A	50p BY105	15p BYZ19	25p OA2	02 7p
ADTI40	50p BC135	30p BC317	12p BF222	80p C760	17p P346A	35p 2N726 17p 2N727	27p BY114 27p BY126	12p OA5 15p OA10	17p SOIG 22p SOIS	
ADZ I I	£2 BC136 £2 10 BC137	30p BC318 35p BC319	12p BF257 12p BF270	35p C762	17p P397	45p 2N743	17p BY 127	17p OA47	7p IN91	4 6p
AFIT4	17p BC139	45p BCY30	12p BF270 20p BF271	25p C764 17p EC401	60p OCP71 15p ORP12	43p 2N744 43p 2N914	17p BY130	15p OA70 35p OA79	7p   IN91 8p   IN41	

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# RADIO FREQUENCIES USED IN SPACECRAFT

A large number of enquiries are received about the communication frequencies used by the Apollo missions and other spacecraft.

The frequencies that are used on spacecraft depend on the matter to be transmitted and also on the propagation conditions. There are normally three bands in use. These are the h.f., v.h.f. and u.h.f. bands. The latter includes the S-band which is of considerable importance since it is less susceptible to propagation anomalies and has perhaps the best noise level characteristic. Noise is at a minimum on this band.

Taking the Apollo spacecraft as an example, there are three units involved. These are the command module, the service module and the lunar module. Communication between ground and the spacecraft from the count down through launch and into earth orbit is maintained by v.h.f. on frequencies of 259MHz and 296MHz (exact values for Apollo 15 were 259.7MHz and 296.8MHz). The command and lunar modules also operate on these frequencies. Within the spacecraft the u.h.f. frequencies are used as well as sound frequencies.

After leaving earth orbit and heading toward the target, communications change to the S-Band. These frequencies are 2273.5MHz for the television channel and 2106.4MHz for speech and exchange of data between the spacecraft and earth. Another channel on 2287.5MHz gives a link between the command module and the ground for speech, directly given data on the course position and other real time information.

The lunar module has two transceivers one of which is used for speech and data and operates on a frequency of 2101.8MHz. The other channel uses a frequency of 2282-5MHz and can be used for data, speech or television. In addition to these systems the craft operates another transceiver which uses a frequency of 10,006MHz. The powers that are used are quite low being 2.8 watts and 11.2 watts on the S-band.

### PARACHUTE TESTING

NASA is to carry out free-flight tests on the Viking landers to test the parachute system to be used in the Viking spacecraft when they are launched in 1975. This is a new departure in technique for no parachute landing of spacecraft has been attempted before.

The Viking system is operated by a mortar which is fired automatically and deploys a 50ft diameter parachute. During tests units will be dropped from 50,000ft and drogue parachutes will slow a ten foot container to the speed at which the mortar will fire.



### PROJECT EOLE

The new French Eole world wide system of weather watch reported previously ran into difficulties.

One hundred and forty-one balloons had been launched in the southern hemisphere and when the satellite was instructed to interrogate it sent instead a destruct signal to 72 balloons which were lost with all their sensors. This occurred on the 364th orbit.

The origin of the wrong signal was traced to one of the control centres at Bretignay. The French space organisation, CNES, consider that the programme will still be valuable since up to 500 balloons will be released during the 180 day period of observation.

### SEARCH FOR EXTRA-TERRESTRIAL LIFE

At the final session of a Conference on Communication there was a discussion for a programme to organise a search for intelligent beings in the space of the galaxy. The conference was attended by many countries and the delegates were from many disciplines: astrophysicists, theoretical astronomers, radio astronomers, sociologists, archaeologists and anthropologists.

# LUNOKHOD

The Russian roving vehicle Lunokhod was shut down after its eleventh day. The last test period resulted in a movement of only 100 metres. It has been shut down because Russian scientists are unwilling to risk the possible misalignment of the on-board laser reflector.

### LAST LUNAR MISSION

The Apollo 17 is, at the moment, the last lunar mission of the series for manned landing on the Moon. The planning of its landing site is therefore of considerable importance. Last June the target decided upon was the area near the crater Alphonsus. It still is the prime candidate.

There are important reasons for the examination of this area. It was from here that the Russian astronomers noted the emmission of gases and changes in the spectrum which suggested vulcanism. Since then, this area has been the subject of close study by professionals and amateurs. Many instances of transient changes in the crater have been noted, some of them detected by spacecraft photographs.

However, after examination of the Apollo 15 data, recovered with much new information, it may be that another site would be of greater scientific value. This sets a problem because the time for crew training is affected particularly as this is the last mission.

### SPACECRAFT COMPUTER

A low power spacecraft computer has been developed by Honeywell. It is suitable for a wide range of unmanned spacecraft missions and its power requirements are only 26.9 watts. It provides a one microsecond memory cycle time.

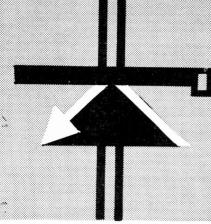
NASA's satellites for application technology will use two of these computers in parallel. The package weighs 23lb and is known as HDC 401.

# SATELLITE GANYMEDE

Ganymede is one of the four satellites originally discovered by Galileo to revolve round Jupiter. It has virtually no atmosphere and it was thought, till recently, that the surface would be like that of the Moon. Recent observations, however, have shown that the atmosphere is below one millibar and the surface may be rock powder or ammonia snow.

These observations were carried out at a wavelength of 25 micrometres at the University of Hawaii. Led by Dr. D. Morrison the team at the University and the Los Alamos Scientific Laboratory observed the satellite before, during and after the eclipse by Jupiter, revealing that except for the presence in the spectrum of features which suggest ammonia frost, the surface is indeed moonlike in character.

The presence in the spectrum of features which suggest ammonia frost still leaves some puzzling questions.



# Substitutes for

By J. N. WATT

HE regulation of fairly low voltages using Zener substitutes is fairly straightforward, if a little unconventional, as was illustrated last month. In this final part we shall look at the application of these substitutes in power supply circuits and give some examples of their applications.

# SHUNT REGULATOR

Fig. 9 shows a typical shunt regulator, where the extra transistor handles the additional power.

The Zener substitute and the power transistor's base-emitter voltage hold the output voltage at  $V_0 = V_{\rm Z} + V_{\rm BE}$ . The resistor R2 ensures that sufficient current flows

through the Zener substitute to give stability.

The current through the shunt transistor is about  $h_{\rm FE}$  times that through the Zener substitute (ignoring the current in R2). It is this that gives the greater power handling capability, which is that of the shunt transistor. This latter could well be one of the unmarked npn devices mentioned earlier as being suitable for use as a Zener substitute.

Calculation of the value of R1 follows the same procedure as in the simple shunt regulator in Fig. 2.

In common with all shunt regulators there are two disadvantages: firstly, stability of output voltage is not as good as can be obtained by other methods, and secondly, the circuit is very inefficient, since full current is taken from the supply whatever the actual load current. With a shunt regulator, the total of load and shunt element currents is a constant.

There is one compensating advantage though—a shunt regulator is inherently short-circuit proof.

### SERIES REGULATOR

For most purposes, series regulators are used when simple low power Zener shunt regulators are not capable of handling the power level required.

Fig. 10 shows the circuit of a typical simple series regulator. The Zener substitute holds the base of the series transistor at a constant voltage and, by emitter follower action in TR1, the output voltage remains constant also, at  $V_0 = V_Z - V_{BE}$ .

Sufficient base current must be supplied to the series transistor TR1 to permit it to pass the required load current. The maximum base current is  $I_{\rm B} = I_{\rm max}/h_{\rm FE}$ , which must be made available by the Zener substitute.

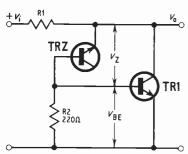
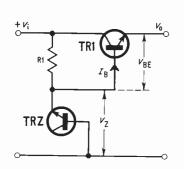


Fig. 9. Transistor shunt regulator



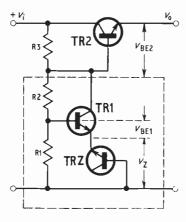
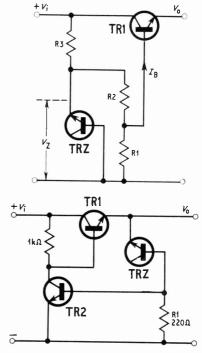


Fig. 10 (left). Simple series regulator

Fig. 11 (above). Series regulator using an "amplified Zener"

Fig. 12 (above right). Division of Zener voltage

Fig. 13 (right). Shunt regulator with improved performance



Practical Electronics December 1971



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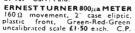
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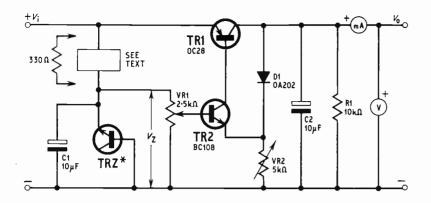


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Fig. 14. Stabiliser with variable output voltage and short circuit pro-

It is as well, therefore, by suitable choice of R1 to arrange for the current through the latter to be, say, 50% greater than this value of  $I_{\rm B}$ .

With no load current, the current taken from the supply is only that through the Zener substitute. However, as it stands the arrangement is unprotected against short circuit of its output, an event which would be most likely to destroy the series transistor, unless protective measures are taken.

### SHUNT TO SERIES CONVERSION

To provide an output voltage greater than  $V_{\rm Z}$ , the circuit in Fig. 6 can be adapted to series operation, as shown in Fig. 11.

$$V_0 = \frac{R_1 + R_2}{R_1} (V_{\rm Z} + V_{\rm BE_1}) - V_{\rm BE_2}$$

An output voltage less than the Zener voltage is easy to arrange by simple division of  $V_{\rm Z}$ .

$$V_0 = \frac{R_1}{R_1 + R_2} V_{\rm Z} - V_{\rm BE}$$

This is shown in Fig. 12.

With these last two circuits sufficient base current must be supplied to permit the series transistor to pass the required full load current.

Thus in Fig. 12, R1 and R2 should be chosen to allow a standing current of, say, five times  $I_{\rm B}$  to flow through them. This current can be looked upon as the load of the simple circuit comprising the Zener substitute and its series resistor R3.

With R1 and R2 so chosen, R3 can be selected as before.

# **VOLTAGE AMPLIFER**

Regulation of the output voltage can be improved by incorporating a further transistor, used as a voltage

amplifier, shown in Fig. 13.

If  $V_0$  tries to rise in level, then so does the voltage at the base of TR2, since the Zener substitute has a constant voltage developed across it. Transistor TR2 amplifies this change, with reversal of sign, so that a larger fall in voltage appears at TR1 base. Emitter follower action in TR1 thus tends to reduce  $V_0$ , so correcting the original rise. TR1 can be a member of the OC28 to OC36 family, while TR2 can be almost any of the popular silicon *npn* devices so readily available.

Besides the better regulation given by the gain of TR2, it is noteworthy that the Zener substitute is driven by the regulated output voltage,  $V_0$ . Since this is constant, better stability of Zener voltage is obtained. In those circuits where the Zener is fed from the unregulated supply, some change of Zener voltage can occur with variation in that supply, due to change of current through the Zener.

An alternative method of obtaining better Zener voltage stability will be mentioned later.

### STABILISED POWER SUPPLY

Having thus briefly described many of the circuits in which Zener substitutes can be employed, we can now turn to the design of a stabilised power supply, using a circuit configuration not previously mentioned, see Fig. 14. It has the following advantages:

- 1. Being a series regulated supply, it can efficiently handle moderately large currents—up to 1A or even greater, depending on the transistor and heat sink employed.
  - 2. It gives a variable stabilised output.
- 3. Despite the series regulator configuration, short circuit protection is provided.
- 4. Constant current output, in place of constant voltage output, is possible, the level of such constant current being easily adjustable.

# SERIES RESISTANCE FOR ZENER

Initially, consider the "black box" (between  $V_i$  and TRZ) feeding the Zener substitute to consist of a resistor of an appropriate value, say 330 ohms. Alternative components for the "black box" will be considered later.

Let us assume that an output current of 1A is required. Then the base current of TR1 will be  $I_{\rm B_1}=1/h_{\rm FE_1}.$  Since this current is provided by TR2, its base current will be  $I_{\rm B_2}=1/(h_{\rm FE_1}h_{\rm FE_2}).$  Assuming  $h_{\rm FE_1}=25$  (OC28 family) and  $h_{\rm FE_2}=100$  (BC108), then the base current of TR2 is  $400\mu\rm A$ .

We should allow, say, 10 times this current to flow through VR1 as a standing current, i.e. 4mA. If  $V_Z$  is made 8 volts (a reasonable value for a Zener substitute), then the value of VR1 is equal  $V_z/4 = 2k\Omega$ .

A  $2.5k\Omega$  potentiometer will be suitable; it does of course give us the means of providing a variable stabilised output.

Capacitor Č1 ensures that the Zener voltage is held smoothly, despite any rapid, large changes in supply voltage  $(V_i)$  which could occur if the mains were used

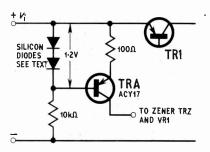


Fig. 15. Constant current drive for Zener substitute

as a primary supply, while C2 reduces any similar changes at the output due to rapid variations in load current. Resistor R1 is necessary to turn on TR1, thereby providing control, by drawing a small load current when no, or only a very small, load current would otherwise be taken. Without R1,  $V_0$  could rise above the stabilised voltage, due to leakage in TR1.

### SHORT CIRCUIT PROTECTION

Turning now to D1 and VR2, it is these components that provide the short circuit protection and constant

current output facility.

In normal running, D1 is forward biased. Should a short circuit be applied to the output of the power supply, it will no longer be so biased, since its anode will then be at negative rail potential (via the short circuit). This results in TR1 passing a current determined by its then existing base current; this will be set largely by the value of VR2, since the lower end of VR2 is now connected directly to the collector of TR1 (via the short circuit).

Thus by choice of an appropriate setting of VR2, limiting of the current to any desired value is obtained.

The constant current mode is brought into play by ensuring that the output voltage setting is such as to attempt to deliver more current than the setting of VR2 allows. Output current is then constant despite supply voltage changes and load changes, provided of course the necessary output voltage can in fact be reached by the supply.

The short circuit protection and constant current output facility are possible in this way because of the method of connection of TR1. It is not, as in previous examples of series stabilised supplies, connected so that emitter follower action takes place. With the output

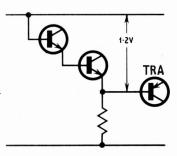


Fig. 16. Use of npn transistors in place of the silicon diodes shown in Fig. 15

taken from the collector as shown, the output impedance is still low, however, because of the heavy negative feedback present.

### CONSTANT SERIES CURRENT FOR ZENER

There is, of course, a disadvantage in this method of connection.

It has been noted above, in connection with the circuit in Fig. 13, that better output voltage stability is obtained if the Zener substitute is driven from the stabilised supply,  $V_0$ . Because of the circuit arrangement, in Fig. 14, of the series transistor being used with its collector as the output, it is not possible to drive the

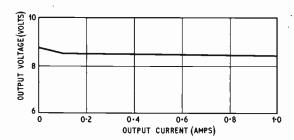


Fig. 17a. Graph of output voltage versus output current for the circuit in Fig. 15

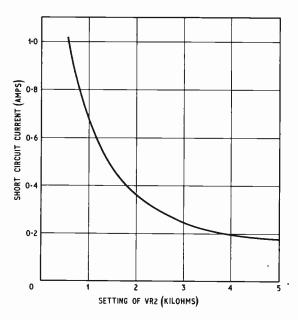
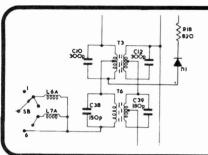


Fig. 17b. Output short circuit current versus setting of VR2

Zener substitute from the stabilised side. Consequently, other means of providing a more constant voltage need to be investigated.

That adopted here concerns the replacement of the "black box" with a constant current circuit similar to that in Fig. 7. The arrangement of this is given in Fig. 15, where a current of 7mA is given by TR1, 4mA through VR1, and 3mA through the Zener substitute.

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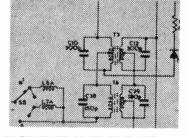


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Arnstrong 525 FM Arnstrong 526 AM/FM Nikko STA501 Nikko STA501 Nikko STA501 Nikko STA501 Nikko STA501 Wharfedale 100-1 New Leak Delta 75 AM FM *New Leak Delta 75 AM FM Arnstrong 524 FM New Leak Delta FM New Leak Delta FM New Leak Delta FM/AM Nikko FAM10 Nikko FAM10 Nikko FAM11	7 u	Tu	mers	MP4.		* * * * * * * * * * * * * * * * * * * *	P. & P. 77 8.8.A. Pri 81- 81- 86- 113- 115- 114- 127- P. & P. 81 8.8.A. Pri 61- 7- 61- 7- 61- 7- 7- 61- 7- 7- 8.8.A. Pri 8.8.A. Pri 8.8.A. Pri 8.8.A. Pri
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Arnstrong 525 FM Arnstrong 526 AM/FM Nikko STA301 Nikko STA301 Nikko STA301 Nikko STA701 Wharfedale 100-1 New Leak Delta 75 AM/FM *New Release.  Arnstrong 523 AM/FM Arnstrong 524 FM Arnstrong 524 FM Arnstrong 524 FM Arnstrong 524 FM Nikko FAM/FM NIKKO	Tu	Tu	mers with dpho	MP4.			P. A. P. 75 S.S.A. Pri
Armstrong 325 FM Armstrong 326 AM/FM Nikko STA301 Nikko STA301 Nikko STA301 Nikko STA701 Wharfedale 190-1 New Leak Delta 75 AM/FM *New Release.  Armstrong 523 AM/FM Armstrong 524 FM Armstrong 524 FM Armstrong 524 FM Armstrong 524 FM New Leak Delta FM *New Leak Delta FM *New Leak Delta FM/AM Nikko FAMI1 Nikko FAMI1 Nikko FAMI1 Nikko FAMI2 Nikko FAMI2 Nikko FAMI2 New Release. Please note: Armstrong 523 Akai ASE9	Tu	Tu	mers with dpho	MP4.			P. A. P. 75 S.S.A. Pri
Armstrong 525 FM Armstrong 526 AM/FM Nikko STA501 Nikko STA501 Nikko STA501 Nikko STA501 Wharfedale 100-1 *New Leak Delta 75 AM/FM *New Release.  Armstrong 523 AM/FM Armstrong 524 FM Armstrong 524 FM Armstrong 524 FM Armstrong 524 FM New Leak Delta FM *New Leak Delta FM/AM Nikko FAMI0 Nikko FAMI0 Nikko FAMI1 Nikko FAMI1 *New Release. Please note: Armstrong 523 A Akai ASE9	Tu	Tu	mers with dpho	MP4.			P. A. P. 75 S.S.A. Pri

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11/21	0.17	AC127	0.25	BF173 BF181	0.90	GJ4M	0.88	0043	0.40
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1N645	0-25	ACY18	0.25	BF195 BF196	0.15	MAT100	0.25	OC46	0-27
1N725A	0-20	ACY19	0.25	BF196	0.15	MATIOI	0.80	OC57	0.60
1N914	0-07	ACVO	0.20	BF197	0.15	MAT120 MAT121 MJE520	0.25	0038	0.60
174914	0.20	ACY20 ACY21	0.20	BF861	0.28	MATIZU	0.20	OC58 OC59	0.00
1N4007	0.20	ACY21 ACY22 ACY27 ACY28 ACY39 ACY40 ACY41 ACY44 AD140	0.20		0.28	MATT21	0.80	OC59	0-65
18021	0.20	ACY22	0-10	BF898	0-28	MJE520	0-87	OC66	0-50
16113	0-15	ACY27	0.25	BFX12 BFX13	0.20	MJE2955	1.37	OC70	0.12
18130	0.18	ACY28	3-17	BFX13	0.25	MJE3055	0-87	OC71	0.12
18131	0.13	ACY39	3-17 0-50		0.25	MJE3055 NKT128 NKT129 NKT211 NKT213 NKT214 NKT216	0.35	0C71 0C72	0.20
18202	0.23	ACV40	0.15	BEX 30	0.25	NKT199	0.30	0073	0.30
2G240	1.98	ACV41	0.15	DEV25	0.98	NETTOIL	0.25	0071	0.30
2G301	0.20	ACV44	0.05	DEVES	0.50	MIL TOIS	0.25	0075	
	0.22	ACI44	0.25	Drago		NK 1213		0075	0.25
2G302		AD140	0.50	BFX84	0.25	NKT214	0.15	OC76	0.25
2G306	0.80	UDITO	0.50	BFX85	0.25	NKT216	0.37	OC77	0-40
2G371	0.22	AD161	0-87	BFX86	0.25	MULTERIA	0.35	0C72 0C73 0C74 0C75 0C76 0C77 0C78 0C79	0-20
2G381	0.25	AD162	0.37	BFX87	0.20	NKT218	0.40	OC79	0.22
2G414	0.80	4 T/10G	0.80	BFX29 BFX30 BFX35 BFX63 BFX84 BFX85 BFX86 BFX87 BFX88 BFY10	0.25	NKT219 NKT222	0.33	OC81 OC81D	0.20
2G417	0.22	AF114	0.25	BFV10	1.00	NK T999	0.20	OC81D	0.20
2N214	0.43	A F115	0.25	BEVII	1.25	NKT224	0.22	OC81M	0.20
2N247	0.25	AF115 AF116	0.25	DYAJI	0.25	NKT251	0.24	OC81DM	0.18
037050	0.50	AP177	0.25	DYDIG	0.25	MKIZDI	0.25	0001031	0.10
2N250	0.00	AF117 AF118	0.20	BFY17 BYF18 BFY19 BFY24 BFY44	0.20	NKT271 NKT272	0.25	OC81Z	0.40
2N404	0.20	AFI18	0.62	BFY19	0.25	NKT272		OC82	0.25
2N697	0.15	AF119	0-20	BF Y24	0-45 1-00	NKT273 NKT274	0.15	OC82D	0.20
2N698	0.40	AF124	0.25	BFY44	1.00	NKT274	0.20	OC83	0.25
2N706	0.10	AF125	0.20	BFY50 BFY51	0-22	NKT275 NKT277	0.25	OC84	0.25
2N706A	0.12	AF126	0.17	BFY51	0.20	NKT277	0.20	OC114	0.38
2N708	0.15	AF127 AF139	0.17	BFY52 BFY53	0.22	NKT278 NKT301	0.25	OC122	0.60
2N709	0.63	AF139	0.30	BFV53	0.17	NKT301	0.40	OC123	0.65
2N711	0.87	AF178	0-55	BEVEL	0.42	NK T204	0.75	00120	0.25
2N987	0.53	AF179	0-45	BFY64 BFY90	0-65	NKT304 NKT403 NKT404 NKT678	0.75	0C139 0C140	0.35
2N1090	0.30	AF180	0-52	DOTES	0.00	NK 1403	0.55	00140	0.80
2N 1090	0.30	AF180		BSX 27 BSX 60	0.50 0.93	NK 1404		OC141 OC169	0.60
2N1091	0.32	AF181	0.42	BSX 60	0.93	NKT678	0.30	OC169	0.20
2N1131	0.25	AF186 AFY19	0.40	BSX76 BSY26	0.15	NKT713 NKT773	0.25	OC170 OC171	0.25
2N1132	0.25	AFY19	1.13	BSY26	0.18	NKT773	0.25	00171	0.30
2N1302	0.18	AFZ11	0.60	BSY27	0.17	NKT777 078B 0A5	0.38	OC200	0.40
2N 1303	0.23	AFZ12	1.00	BSY51	0.50	O78B	0.38	OC901	0.70
2N1304	0.25	A8Y26 A8Y27	0.25	BSY95A	0.12	0.45	0.20	OC202 OC203	0.80
2N1305	0.22	ASVOT	0.32	BSY95	0.12		0.12	00202	0.40
2N1306	0.25	AGVON	0.02	BT102/50	0.15	0A47 0A70	0.10	OC203	0.40
2N1300	0.25	ASY28 ASY29	0.25 0.30	B T 102/50	0.75	0A47	0-10	0C204 0C205	
2N1307	0.50	ASY29 ASY36 ASY50 ASY51 ASY53 ASY55 ASY62 ASY86	0.30		0.75	OA70	0.10	OC205	0.75
2N1308	0.25	ASY36	0.25	BTY42 BTY79/10	0.92		0.10	OC206	0.90
2N1309 2N1420	0.25	ASY50	0-17	BTY79/10	00R	0A73 0A74 0A79	0.10	OC207	0.90
2N1420	0.93	ASY51	0.40	l	0.75	0A74	0.10	OC460 OC470	0.20
2N1507	0.28	A8Y53	0.20	BTY79/46	00R	OA79	0.10	OC470	0.80
2N1526	0.38	A8Y55	0.20		1.25	0A81	0.08	OCP71	0-97
2N1909	0-38 2-25	ASV69	0.25	BY100	0.15	0A85	0.12	OCP71 ORP12	0.50
2N2147	0.75	19V86	0.83	DITTOO		OA86	0.15	ORP60	0-40
2N2148	0.60	ASY86 ASZ21 ASZ23 AUY10 AU101 BC107 BC108 BC109 BC113 BC115		BY126	0.15	0.400	0.08	ORP61	0.42
2N2148 2N2160	0.00	A5221	0.42	BY127	0.17	OA90	0.08		
2N2160	0.60	A8223	0.75	BY182	0.85	OA91	0.07	S19T	0.80
2N2218	0.20	AUY10	0-98			OA95	0.07	SAC40	0-25
2N2219	0.20	AU101	1.50	BY213	0.25	OA200	0.07	SFT308	0.38
2N2287	1.03	BC107	0.10	BYZ10	0.35	OA200 OA202 OA210	0.10	SAC40 SFT308 ST722 ST7231	0.38
2N2297	0.20	BC108	0-10	BYZII	0.32	OA210	0.25	ST7231	0.68
2N2369A	0-15	BC109	0.10			0A211 0AZ200	0.30	SX 68 SX 631	0.20
2N2613	0.28	BCI13	0.15	BYZ12	0.80	0 4 2 2 0 0	0.55	8 Y 631	0.30
2N2646	0.45	BC115	0.20	BYZ13	0.25	0.12200	0.50	SX 635	0-40
03/0710	0.25	BC116	0.20	BYZ15	1.00	OAZ201 OAZ202	0.40	0 V 0 10	0.40
2N2712	0.20	DCIII	0.25			UAZ202	0.40	SX 640	0.50
2N2784 2N2846	0.50	BC116A	0.30	BYZ16	0.62	OAZ203 OAZ204	0.42	SX641	0.55
2N2846	0.75	BC118	0.25	BYZ88C3	V3	OAZ204	0.30	SX 642	0.60
2N2848	0.42	BC121 BC122	0-20		0.15	OAZ205	0.42	SX 644	0.75
2N2904	0.20	BC122	0.20	CHI	0.65	OAZ205 OAZ206	0.42	SX645	0.75
2N2904A	0-38	BC125	0-68	CRS1/05	0.25	OAZ207	0.47	V15/30P	0.50
2N2906	0.20	BC126	0.65	CRS1/40	0.25 0.17	OAZ208 OAZ209	0.32	V15/30P V30/201P	0.75
2N2907	0.28	BC140	0-55	CS4B	2.50	OAZ209	0.83	V60/201	0.75
2N2924	0.23	BC147	0.15	CS10B	3.13	OAZ210	0.32	V60/201P	0-38
2N2925	0.15	BC148	0.12	DD000	0.15	0AZ211	0.82	V A 101	0.10
2N2926	0.10			DD000	0.15	OAZ222	0.45	X A101 X A102 X A151	0.18
2N 2920 2N 3054	0.50	BC149	0.20		0.18	OAZ222 OAZ223	0.45	VALUE	
211 3004 031 0077	0-50	BC157	0.15	DD006 DD007	0.18	0.47004	0.45	AAIDI	0-15
2N3055	0.75	DOLLE		DD007	0.40	OAZ224	0.45	XA152	0.15
2N3702	0.10	BC158	0.12	DD008	0-38		0.22	X A 161	0.25
2N3705	0.10	BC160	0.63	GD3	0.33	OAZ242	0.23		
2N3706	0.23	BC169	0.12	GD4	0.05	OAZ241 OAZ242 OAZ244 OAZ246 OAZ290 OC16	0.23	XA162	0.25
2N3707	0.12		0.30	GD5	0.33	OAZ246	0.23	XB101	0.48
2N3709	0.13	BCY31		GD8	0.25	OAZ290	0.38	XB102	0.10
2N3710	0.13	BCY32	0.55	GD19	0.05	OCIA	0.50	XB102	
2N3711	0.10	TO CATE OR	0.25	CETION	0.80	OCIET	0.38		0.25
2N3711 2N3819	0.35	BCY34 BCY38 BCY39 BCY40 BCY42 BCY70 BCY71 BCZ10 BCZ11 BD121	0.30	GET102 GET103	0.22	OC16T OC19	0.38	X B113	0.12
2113019	0.99	DC 134	0.30	GET103		0019		XB121	0.48
2N3820	0-60	BCY38	0.40	GET113 GET114 GET115 GET116 GET120	0.20	OC20 OC22	0.98	ZR24	0.68
2N3823	0.75 0.58	BCY39	1.00	GET114	0.15	OC22	0.50		
2N5027	0.53	BCY40	0.50	GET115	0.45	OC23	0-60	ZS170	0.10
2N5088	0.33	BCY42	0.25	GET116	0.50	OC24	0.60	Z8271	0 18
28005	1.00	BCY70	0.15	GET120	0.25				
28301	0.50	BCY71	0.20	GET879	0.30	OC25	0.37	ZT21	0.25
28304	0.25	BCZ10	0.35	GET872 GET875	0.25	OC26	0.25	ZT43	0.25
28501	0.75	BCZ11	0.50	GET880	0.87	OC28	0.60	ZTX107	0.15
		D D 191	0.65	CETOOU		OC29			
28703	0.62	BD121		GET881	0.25	0029	0.60	ZTX108	0.12
AA129	0.20	BD123 BD124	0.80	GET882	0.25	OC30	0-40	ZTX300	0.12
AAZ12	0.30	BD124	0.75	GET885	0.25	OC35	0.50	ZTX304	0.25
AAZ13	0.12	BDYII	1.62	GEX44	0.08	OC36	0.80	2TV 500	
	0.1%	BF115 BF117	0.25	GEX45/1 GEX941	0.07		0.00	ZTX500	0.16
AC107	0.37	BF117	0.50	GEX 941	0.15	OC41	0.25	ZTX503	0.17
AC126	0.20	BF167	0.25	GJ3M	0.25	OC42	0.30	ZTX531	0.25
			- ~-						

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Terms C.W.O. only Tel. 01-769 0199/1649 A modification of the original constant current circuit is employed, due to the need for it to function correctly with a varying supply voltage,  $V_i$ . The two silicon diodes (these can be of any low power type, or alternatively two npn silicon transistor base-emitter junctions, forward biased as shown in Fig. 16) provide a constant voltage at the base of TRA—constant, that is, with respect to the positive supply line. With the base and hence emitter so held constant, the 100 ohm resistor passes a constant 7mA current through TRA and hence to the TRZ and VR1 in parallel.

It was from this configuration that the graphs in Fig. 17 were obtained, to show how output voltage varies with output current, and how the level of short circuit current varies with the setting of VR2.

In passing, note that the driving of the TRZ from the input voltage allows the output voltage to be made variable down to zero volts. If TRZ is driven from the output side, output voltages of less than Zener voltage would cease to give regulation.

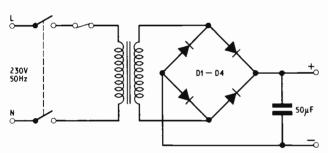


Fig. 18. Suggested mains supply unit for the stabiliser shown in Fig. 14

## **VOLTAGE FEEDER**

Practical construction of such a stabilised power supply can follow usual techniques. After maximum output current and maximum output voltage have been decided upon, a decision is made on the method of driving the regulator. A simple mains unit is shown in Fig. 18, and the voltage rating of the transformer secondary is chosen to give, at full load current, a value of  $V_i$ , about 4 volts more than the maximum regulated output voltage. Alternatively, a battery or accumulator could provide power.

### **HEAT SINK FOR TRI**

As for the size of the heat sink required for TR1, calculate first the maximum power dissipated in that transistor. This will be:

 $V_i$  – (minimum output voltage used)]

× maximum load current

It is at minimum output voltage and maximum load current that maximum power will be dissipated in TR1, and a suitable heat-sink should be chosen.

Generally speaking, for an OC28 or similar transistor, bolted to a heat-sink with an insulating mica washer, a rise in heat-sink temperature of about 40°C above ambient room temperature can be permitted.

Heat-sinks of various °C/watt ratings are readily available from components stockists.

As an example, for a 1A 9 volts output, driven from a 13 volts supply, with operation down to 3 volts expected, maximum power dissipated in TR1 is  $(13-3) \times 1 = 10$  watts, so a heat-sink of 4°C/watt is needed for an OC28. About 20 sq in of 16 s.w.g. aluminium sheet will suffice.

Having thus decided on transformer and heat-sink requirements, the size of the final unit can be settled. VR1, and possibly VR2, are front panel controls, but otherwise layout is unimportant, and can be adapted to suit individual requirements.

# TRICKLE CHARGER FOR SMALL BATTERIES

Mention was made earlier of deriving the supply from an accumulator and this is a useful way of obtaining a good general purpose stabilised voltage for powering many circuits. Recharging of the accumulator can be carried out overnight or a trickle charger can be used while the power supply is in use.

It is worth noting that many 12 volt car batteries can still provide a handy 1A even after their useful life in a car is over.

Constant current output can prove useful in the charging of some types of small batteries.

Batteries such as nickel-cadmium have a low internal resistance; hence, when in a discharged state, they would draw very large currents if the usual constant (or almost constant) voltage charging system is employed. This large current could give rise to overheating, tending to reduce the internal resistance still further and hence allowing thermal runaway to take place, and so leading to destruction of the battery.

However, for battery charging applications, adjust VR2 to give the required current, ensuring that VR1 is set to give a high enough voltage to charge the battery in question when it reaches its full final voltage. The battery manufacturer's instructions should be followed if in doubt.

Otherwise, the constant current circuitry is used as an overload limiter. With VR2 at maximum resistance, short circuit the output terminals; then bring down VR2 until the output current is a little greater than the maximum expected to flow in the unit being powered, under normal conditions.

Thereafter, should excessive current attempt to flow due to a fault, such current will be safely limited to that previously set by VR2.

Although the uses of Zener substitutes so far described have been in power supplies, this does not exhaust their usefulness.

# **OTHER USES**

Clipping and limiting of waveforms to specific amplitudes can easily be accomplished. In the design of d.c. amplifiers, shifts of voltage by the constant voltage of the Zener substitute enable large gains to be achieved without limiting due to saturation of later stages.

With the ready availability of *npn* silicon transistors for use as substitute Zeners, it is hoped that experimenters will be encouraged to make greater use of them in stabilised power supplies, on the lines suggested, with corresponding improvement in equipment performance.

# MARKET PLACE

Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

# PHOTOCONDUCTIVE CELL DESIGNERS' KIT

Designed primarily for educational establishments as well as industrial organisations interested in developing light activated controls, **Photain Controls** have produced a Photoconductive Cell Designers' Kit which readers may like to experiment with.

The kit consists of five different photoconductive cells together with 13 circuit diagrams each complete with a brief explanation of the circuit. Each diagram enables the student to build devices embracing light intensity measurement; feedback gain controllers; photo switches; signal modulating and musical instrument volume controller.

The kit is available from Photain Controls Ltd., Randalls Road, Leatherhead, Surrey, price £2.

### SOLDERING ACCESSORIES

Leaving one hand free to hold the work piece, the Anextra Solder Feed fits the majority of soldering irons and carries up to 4oz reels of flux cored solder.

Solder from 22 to 18 s.w.g. can be used and is fed to the joint by operating the solder feed trigger.

Supplied with an initial 10z reel of 60/40 cored 22 s.w.g. solder, the recommended retail price is £4:25. Further details of the Solder Feed is obtainable from Anextra Ltd.,

Chiltern Works, 77-78 Chiltern View Road, Uxbridge, Middlesex.

Standard package integrated circuits, despite their numerous advantages, suffer from one major drawback as far as the amateur is concerned. The removal of these units from printed circuit boards, since it is necessary either to remove all traces of solder from all the connecting pin joints or melt all the joints simultaneously.

A range of desoldering heads for the Solderstat HMS miniature irons is an accessory which, using the method of simultaneous desoldering, removes the standard dual-in-line packages within a few seconds is claimed by Solderstat Ltd.

The desolder head is placed on the iron in place of the copper bit and aligned with the i.c. connecting pins on the printed circuit board. Both 14-way and 16-way dual-inline heads are available.

Readers can, of course, use standard i.c. holders available from advertisers when they construct prototype circuits.

# LOW VOLTAGE TOOLS

A range of battery-operated miniature tools capable of operating drills, cleaning brushes, abrasive stones, cutting burrs, polishing mops and other tools for precision work has been introduced by Expo (Drills) Ltd.

Two basic models are available. The "Reliant" designed for lighter work such as model making, and has a full load current of 1.5A, and the "Titan Super". Rated torque of the "Reliant" is 1.38oz in (100gm cm.).

For jobs requiring a more powerful tool and for professional applications the "Titan Super", rated at 3.5A on full load, should be used. It has a rated torque of 350 c.m.p. operating at 4,000-9,000 r.p.m.

Accessories include a diamond bonded drill, for gem stones up to 7MOHS, various types of abrasives, cutters and saws. Different collets and accessories are supplied according to model.

Prices range from £3 to £5.50 and full details are obtainable from Expo (Drills) Ltd., 62 Neal Street, London, W.C.2.

### INSTANT LETTERING

To help finish-off equipment Instant Plastic Ltd. have introduced a range of p.v.c. self-adhesive plastic lettering kits.

Designated type K, the kits consist of 40 different sets in five colours, sizes from 10mm (\$\frac{1}{4}\$in) high upwards, in capitals and lower case letters and numerals.

The smallest set contains 578 letters and numerals and the number in each set reduces as the letter size increases.

The K sets cost approximately £1 per set, and details of their other ranges can be obtained from Instant Plastic Ltd., 101 Bramley Road. London, W.10.

# PRINTED CIRCUIT BOARDS

A new price list of printed circuit boards for previously published P.E. projects has just been issued by P.H. Electronics.

Copies of the price list can be obtained from P.H. Electronics, Industrial Estate, Sandwich, Kent.

### SWITCH KIT

The mention of the versatile switch kit in the September issue should have included the following firms who are all part of the buying group set up by Home Radio (Components Ltd.).

Crescent Radio, 40 Mayes Road, London, N.22.

Garland Bros., Chesham House, Deptford Broadway, London, S.E.8.

Radioparts, 5 Market Way, Plymouth, Devon.

Servio Radio, 156-158 Merton Road, Wimbledon, London, S.W.19.



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A.		500
District Co.		1m2
		300
50μA	£3-60	1A
50.0.50μA	£3-45	5.4
10044	£9.90	300

YPE SW. 100 00 x 80 mm £3·35 £3·10 £3·10 £3·10 £3·10 d.c. . . . V d.e. £3·10 | 1mA ... £3·10 | 20V d.e. £3·10 | 50V d.e. £3·75 | 300V d.e. 1A d.c. 5A d.c. £3.45 £8.20 £3.45 300V a.c. VU Meter

TYPE S-80 80 mm square fronts £3.20 £3.10 £2.75 £3.10 50μA ..... 50-0-50μA .. 100μA . . . . 100κ-0-100μΔ 500µA .. £3.00 £2-60 £2.80



,	AC PERSONS	Contract to
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)	5A d.e	£2-60
)	300V a.c	£2.60
)	VU Meter	£3.37

### "SEW" CLEAR PLASTIC METERS 41in 41in fronts.

Type mr. oor. *	\$1H #\$1H
	50mA
DANGES EXCHES FOR MORE 19	100mA
S meture 5	500 mA
	; IA
m.A	5A
01	15A
and a second	30A
	20 V d.c
	50V d.c

100μΑ ..... 100-0-100μΑ

, 1	3	500mA	£2-80
	the same	1A	£2.80
mA	Y	5A	£2-80
07		15A	£2.80
100 10 mm	***	30A	£2.90
		20V d.c	£2.80
4	-	50V d.c	£2.80
~		150V d.c	£2.80
50μA	£3-60	300V d.c	£2.80
50-0-50μA	£3.10	15V a.c.	£2.80
100μΑ	£3.10	300V a.c	£2.80
100-0-100µA	£3.00	S Meter	***
200μΛ	£3.00	1mA	£2.87
500μA	£2.90	VU Meter	£3-60
$500 \text{-} 0 \text{-} 500 \mu \text{A}$	£2.80	1A a.c.*	£2.80
ImA	£2-80	5A a.c.*	£2.80
1-0-1mA	£2.80	10A a.c.*	£2.80
5mA	£2.80	20A a.c.*	£2.80
$10\mathrm{mA} \cdot \ldots \cdot$	£2.80	30A a.c.*	£2.80

omA	£2.80	20A a.c.*	£2.80
10mA	£2.80	30A a.c.*	£2.80
Type MR	.52P. 2i	in square fronts	1.
$50\mu A$	£3·10	1 20V d.e	£2.00
50-0-50μA	£2.60	50V d.e	£2.00
100μΑ	£2.60	300V d.c	£2.00
100-0-100µA	£2.50	15V a.e	£2.10
500μΑ	£2.30	300V a.c	£2-10
1mA	£2.00	S Meter	
5mA	£2.00	1mA	£2·10
10mA	£2.00	VU Meter	£3.20
50mA	£2-00	1A a.c.*	£2-00
100mA	£2.00	5.A a.c.*	£2.00
500mA	£2.00		£2.00
1A	£2.00	10A a.c.*	
5A	£2.00	20A a.c.*	£2.00
10V d.c	£2.00	1 30A a.e.*	£2-00

201 4.6	22.00	more with the	22 00
Type MR	.65P. 3	in . 3 in fronts	١,
50µA	£3-37	20V d.e	£2-20
50-0-50μA	£2.75	50V d.c	£2-20
100µA	£2.75	150V d.c	£2-20
100-0-100µA	£2.65	300V d.c	£2-20
200μΑ	£2.65	15V a.c	£2.30
500μA	£2-40	50V a.c	£2-30
500·0-500µA	£2-20	150V a.c	£2.30
ImA	£2-20	300V a.c	£2.30
5mA	£2.20	500V a.c.	£2.30
10mA	£2-20	S Meter	
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100mA	£2-20	VU meter	£3-37
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1A	£2-20	100mA a.c.*	£2-20
5A	£2.20	200mA a.c.*	£2.20
10A	£2.20	500mA a.c.*	£2-20
154	£2-20	IA a.c.*	£2-20
20A	£2-20	5.4 a.c.*	£2.20
30A	£2.30		
50A	£2.50	10A a.c.*	£2-20
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n the followin	g ranges		
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100μA		50V d.c £4.40	PE.70
$50-0-50\mu A$	£4-65	300V d.c £4-40	
1-0-1mA		Dual range	50μΛ . 50-0-50
1A d.c		500mA/ŠA d.c. <b>£4-65</b>	100µA 100-0-1
10V d.c	£4-40	5V/50V d.c. £4.65	200 μ.Δ

# Type MR.38P. 1 21/32in square fronts

in 41in ironts.		Type mm.35P. 1 21/	ONLY BELLE TO	
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	£2.80	1	200mA	£1-60
500mA	£2-80	annuals 2.01 MOSc not man visit	300mA	£1.60
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	£2.80	. <b>48</b> -1	10A	£1.60
	£2.80	· '	3V d.e	£1.60
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	£2.87	100-0-100µA £1.75	150V d.c	£1-60
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1A a.c.*	£2.80	500μA £1-65	500V d.e	£1.60
5A a.c.*	£2.80	500-0-500μA £1-60	750V d.e	£1.60
10A a.c.*	£2.80	ImA £1.60	15V a.e	£1.70
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30A a.c.*	£2.80	2mA £1.60	150V a.c.	£1.70
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	£2.00	20mA £1-60	8 Meter	
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50mA	£1.70	5A a.e.*	£1.7
100mA	£1.70	10A a.c.*	£1.7
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100-0-100μA 500μA . . . . 1mA . . . . . 1-0-1mA . . . .

5mA

10mA

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		20V d.c	£1.90
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	£2-25	500mA a.c.*	£1.9
	£2-20	1A a.c.*	£1.9
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10. 50, 250, 100.1,000V.
D.c. current: 10, 100µA, 10, 100,000
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Should any item be out of stock we reserve the right to supply at higher voltage rated item at no extra charge.

TERMS: Cash with order please. Postage: 10p inland, 25p Europe. Elsewhere — send plenty — will refund excess.
All goods carry manufacturers warranty.
Counter Sales—Same Address.

X 57	Mulia 250v.	rd	Me	ta	llise	d		ester series
E	0.01							3р
В	0.015							3p 3p
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400	16	12
640	16	15
1.000	16	18
160	25	9
250	25	12
400	25	15
640	25	18
100	40	9
160	40	12
250	40	15
400	40	18

Mullard Sub-Miniature Ceramic Plate C333 series 63 volt working. Range | 8pf to Plate C. 63 volt working. Range 220pf (usual pref. values). Packs of 6 (any values) 150

	ITORS-Mul	lard Minia
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Mfd.	Volt Wkg.	
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		LINEAR	AND DI	GITA	L ICs		
R.C.A.			Fairchild		1-11	12-24	25 +
CA3004		£1-80	µL900		40p	35p	32p
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CA3011		75p	μL923		53p	50p	47p
CA3013		£1-05	Devices	may	be mixed		qualify for
CA3014		£1-25			quantity pri	ce.	
			G.E. (U.				
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CA3020		£1:30	PA234		itt Amp		£1-00
CA3028A		75p	PA237		itt Amp		£1-87
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CA3043		£1-40	PA424	Zero	o Volt Switc	n	12.43
CA3044		£1-20	Mullard				75.
CA3046		75p	TAA263		ar Amp.		75p
CA3047		£1 40	TAA293 TAA310		. Purp. Amp ord/Playback		. £1.50
CA3048		€2:05	TAA320		S LF Amp	Amp	65p
CA3049		£1:60	TADIO		leceiver		£1.97
CA3052		£1:65	TADITO		FM Receiver		€1.97
CA3090Q		£3 ·46	TADITO	A1 1)	TIT RECEIVE		
BARGAIN							
OP-AMPS!!		1.0	5				
LM709C		60p		USI	D REC	111	PEDII
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op-am	p)	05-	P.	LESSI	EY IC AMP	C	.50
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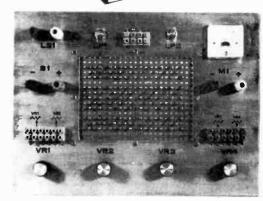




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35 35 35 50 50 35 75 20 22 27 56 56 150 0·18 µF 0·33 µF 0·47 µF 0·68 µF 1·0 µF 2·2 µF 2·7 µF 2·7 µF 3·0 µF 10·0 µF volts volts volts volts 0.47 0.68 µF 0.68 µF 1.0 µF 1.8 µF 2.2 µF 2.7 µF 3.3 µF 4.7 µF 5.6 volts volts volts volts 3 volts 15 volts 35 volts 12 volts 1.5 volts volts 50 20 150 35 50 20 20 15 6 voits volts 20 50 12 15 volts volts volts volts volts volts voits volts

We have huge numbers of components in quantities too small to advertise individually. In order to "clear the decks" we have made up parcels containing a mixture of carbon and wire-wound resistors, electrolytic and paper condensers, controls, transistors, diodes etc., for a tiny fraction of normal price, it is emphasised that these are mixed parcels only—contents cannot be stipulated! Sold only by weight.

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### NEW! NEW! NEW! NEW!

An aerosol spray providing a convenient means of producing any number of copies of a printed circuit both simply and quickly.

Method: Spray copper laminate board with light-sensitive Cover with transparent film upon which circuit has spray. been drawn. Expose to light. (No need to use ultra-violet.) Spray with developer, rinse and etch in normal manner,

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(4) Offer closes January 31, 1972.

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Pye Transistorised Car Radio Panels includes 2 × Double Tuned IFs, 5 NKT Transistors, etc., etc. Your bargain for 50p, post 10p U.K.
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For other bargains see our advert on page 542 October Practical Wireless, Our location near Goodge St. Station, opposite Heals of Tottenham Court Road.

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RADIO INTERFERENCE SUPPRESSED
For all petrol engines—cars, boats, etc. Guaranteed for 5 yrs.



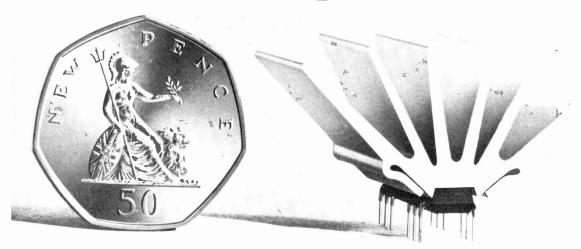
Complete Installation Kit for I2-volt vehicles £12-95 + 35p P, & P. State earth polarity of vehicle POSITIVE or NEGATIVE earth. Unit Construction Kic also available for the radio electronics constructor £9-95 + 35p P, & P. The construction kic includes instructions and all components for wiring as positive or negative earth, and is complete with the stove enamelled steel case and aluminium base. All components are available separately.

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# Super IC-12



# High fidelity Monolithic Integrated Circuit Amplifier

Two years ago Sinclair Radionics announced the World's first monolithic integrated circuit Hi-Fi amplifier, the IC.10. Now we are delighted to be able to introduce its successor, the Super IC.12. This 22 transistor unit has all the virtues of the original IC.10 plus the following advantages:

- 1. Higher power.
- Fewer external components.
- 3. Lower quiescent consumption.
- 4. Compatible with Project 60 modules.
- Specially designed built-in heat sink. No other heat sink needed.
- 6. Full output into 3, 4, 5 or 8 ohms.
- 7. Works on any voltage from 6 to 28 volts without adjustment.
- 8. NEW 22 transistor circuit.

# SINCLAIR GENERAL GUARANTEE

Should you not be completely satisfield with your purchase when you receive it from us, return the goods without delay and your money will be refunded in full, including cost of return postage, at once and without question. Full service facilities are available to all Sinciair customers.

Output power 6 watts RMS continuous (12 watts peak).

Frequency Response 5 Hz to 100KHz ±

**Total Harmonic Distortion** Less than 1%. (Typical 0.1%) at all output powers and all frequencies in the audio band.

Load Impedance 3 to 15 ohms.

Power Gain 90dB (1,000,000,000 times) after feedback.

**Supply Voltage** 6 to 28 volts (Sinclair PZ-5 or PZ-6 power supplies ideal).

Size 22 x 45 x 28 mm including pins and heat sink

Input Impedance 250 Kohms nominal.

Quiescent current 8mA at 28 volts

With the addition of only a very few external resistors and capacitors the Super IC.12 makes a complete high fidelity audio amplifier suitable for use with pick-up, F.M. tuner etc. Alternatively, for more elaborate systems, modules in the Project-60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC.12 ideal for battery operation.



Price, inc. FREE printed circuit board for mounting.

£2.98

Sinclair Radionics Ltd, London Rd, St. Ives Huntingdonshire PE17 4HJ Telephone St Ives (048 06) 4311



Practical Electronics December 1971

# Sinclair Project 60

# The World's leading range of high fidelity modules







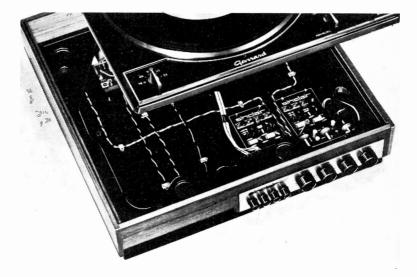
Project 605 is one pack containing: one PZ5, two Z30's, one Stereo 60 and one Masterlink. This new module contains all the input sockets and output components needed together with all necessary leads cut to length and fitted with neat little clips to plug straight on to the modules. Thus all soldering and hunting for the odd part is eliminated. You will be able to add further Project 60 modules as they become available adapted to the Project 605 method of connecting.

Complete Project 605 pack with comprehensive manual, post free £29.95
All you need for a superb 30 watt high fidelity

Sinclair Radionics Limited, London Road, St. Ives. Huntingdonshire PE17 4HJ. Tel: St. Ives (048 06) 4311

stereo amplifier.





Project 60 offers more advantage to the constructor and user of high fidelity equipment than any other system in the world.

Performance characteristics are so good they hold their own with any other available system irrespective of price or size.

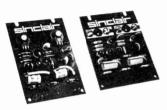
Project 60 modules are more versatile — using them you can have anything from a simple record player or car radio amplifier to a sophisticated and powerful stereo tuner-amplifier. Either power amplifier can be used in a wide variety of applications as well as high fidelity. The Stereo 60 pre-amplifier control unit may also be used with any other power amplifier system, as can the AFU filter unit. The stereo FM tuner operates on the unique phase lock loop principle to provide the best ever standards of sensitivity and audio quality. Project 60 modules are very easily connected together by following the 48 page manual supplied free with all Project 60 equipment. The modules are great space savers too and are sold individually boxed in distinctive white and black cartons. With all these wonderful advantages, there remains the most attractive of all — price. When you choose Project 60 you know you are going to get the best high fidelity in the world, yet thanks to Sinclair's vast manufacturing resources (the largest in Europe) prices are fantastically low and everything you buy is covered by the famous Sinclair guarantee of reliability and satisfaction.

# Typical Project 60 applications

The Units to use	together with	Cost of Units
Z.30	Crystal P.U., 12V battery volume control	£4.48
Z.30, PZ.5	Crystal or ceramic P.U. volume control etc.	£9.45
2 x Z.30s, Stereo 60, PZ.5	Crystal, ceramic or mag. P.U., F.M. Tuner, etc.	£23.90
2 x Z.30s, Stereo 60, PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc.	£26.90
2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr	As above	£34.88
Z.50, PZ.8, mains transformer	Mic., guitar, speakers, etc., controls	£19.43
	Z.30, PZ.5  2 x Z.30s, Stereo 60, PZ.5  2 x Z.30s, Stereo 60, PZ.6  2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr  Z.50, PZ.8, mains	Z.30 Crystal P.U., 12V battery volume control  Z.30, PZ.5 Crystal or ceramic P.U. volume control etc.  2 x Z.30s, Stereo 60, PZ.6 P.U., F.M. Tuner, etc.  2 x Z.30s, Stereo 60, PZ.6 As above  2 x Z.50s, Stereo 60 PZ.8, mains trsfrmr  Z.50, PZ.8, mains Mic., guitar, speakers, etc.,

# from a simple amplifier to a complete stereo tuner amplifier with Project 60 modules

# Z.30 & Z.50 power amplifiers



The Z.30 and Z.50 are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at full output and all lower outputs. Whether you use Z.30 or Z.50 amplifiers in your Project 60 system will depend on personal preference, but they are the same size and may be used with other units in the Project 60 range equally well.

SPECIFICATIONS (2.50 units are interchangeable with Z.30s in all applications). Power Outputs

Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M.S. into 3 ohms using 30 volts. **Z.50** 40 watts R.M.S. into 3 ohms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.
Frequency response: 30 to 300,000Hz + 1dB

Distortion: 0.02% into 8 ohms.

Signal to noise ratio: better than 70dB unweighted.

Input sensitivity: 250mV into 100 Kohms For speakers from 3 to 15 ohms impedance. Size: 14 x 80 x 57 mm

Z.30

Built, tested and guaranteed with circuits and instructions manual. £4.48

Built, tested and guaranteed with circuits and instructions manual. £5.48

# Power Supply

Designed special for use with the Project 60 system of your choice. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ.5 30 volts unstabilised £4.98 PZ.635 volts stabilised £7.98 PZ.8 45 volts stabilised (less mains transformer) £7.98 PZ.8 mains transformer £5.98

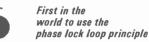


# The Sinclair Guarantee

If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.

# Project 60 Stereo F.M. Tuner





The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other original features include varicab diode tuning. printed circuit coils, an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Good reception is possible in difficult areas, and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high ficelity system.

SPECIFICATIONS-Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108 MHz. Capture ratio: 1.5d8. Sensitivity: 2μV for 30d8 quelting: 7μV for lock-in over full deviation. Squelch level: 20μV. A.F.C. range: ±200 KHz. Signal to noise ratio: >65d8. Audio frequency response: 10 Hz - 15 KHz (±1d8). Total harmonic distortion: 0.15% for 30% modulation, Stereo decoder operating level:  $2\mu V$ . Cross talk: 40dB. Output voltage:  $2 \times 150 \text{mV}$  R.M.S. Operating voltage: 25-30 VDC. Indicators: Power on/tuning/stereo. Size: 93 x 40 x 207 mm

Built and tested. Post free.

# Stereo 60 Pre-amp/control unit



Designed for Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

means of push buttons and accurate equalisation is provided for an one assumingsto.

SPECIFICATIONS—Input sensitivities: Radio – up to 3mV. Mag. p.u. 3mV; correct to R.I.A.A curve ±1dB:20 to 25,000 Hz. Ceramic p.u. – up to 3mV: Aux – up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE ± 15 to —15dB at 10 KHz: BASS ± 15 to —15dB at 100Hz.

Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm.

£9.98

Built tested and guaranteed.

# A.F.U. High & Low Pass Filter Unit



For use between Stereo 60 unit and two Z.30s or Z.50s. and is easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less

loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. Two filter stages – rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current – 3mA. H.F. cut-off (–3dB) variable from 28KHz to 5KHz. L.F. cut-off (–3dB) variable from 25Hz to 100Hz. Distortion at 1KHz (35V, supply (0.02% at rated FE QQ output. Size: 66 x 40 x 90 mm. Built tested and guaranteed.

To: SINCLAIR RADIONICS LTD LONDON ROAD	ST. IVES HUNTINGDONSHIRE PE17 4HJ
Please send	Name
	Address
l enclose cash/cheque/money order.	P.E. 1271

Practical Electronics December 1971

# Sinclair Q16/Micromatic

### Q16 High fidelity loudspeaker

The Q16 employs the well proven acoustic principles specially developed by Sinclair in which a special driver assembly is meticulously matched to the characteristics. of the uniquely designed cabinet. In reviewing this exclusive Sinclair design, technical journals have justly compared the Q16 with much more expensive loudspeakers. Its shape enables the Q16 to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures. A solid teak surround with a special all-over cellular foam front is used as much for appearance as its ability to pass all audio frequencies without loss.

This elegantly designed shelf mounting speaker brings genuine high fidelity within reach of every music lover.

### Specifications:

Construction: Special sealed seamless sound or pressure chamber with internal

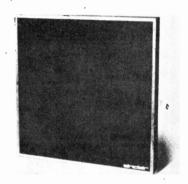
Loading: up to 14 watts RMS. Input Impedance: 8 ohms.

Frequency response: From 60 to 16,000 Hz. confirmed by independently plotted B and K curve

Driver unit: Special high compliance unit having massive ceramic magnet of 11,000 gauss, aluminium speech coil and special cone suspension for excellent transient

Size and styling: 93 in, square on face x 43 in, deep with neat pedestal base. Black all over cellular foam front with natural solid teak surround

Price £8.98.



### Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage AM receiver brilliantly designed to provide remarkable standards of selectivity, power and quality for its size. Powerful AGC counteracts fading from distant stations: bandspread at higher frequencies makes reception of Radio 1 easy. The plug-in magnetic earpiece provided, matches the Micromatic's output to give wonderful standards of reproduction. Everything including the special ferrite rod aerial and batteries is contained within the minute attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built and tested, you will find it as easy to take with you as your wrist watch, and dependable under the severest listening conditions.

### Specifications:

**Size:**  $36 \times 33 \times 13 \text{ mm}$  (1.8 x 1.3 x 0.5 in.) Weight: including batteries, 28.4 gm

Case: Black plastic with anodised aluminium front panel and spun aluminium

Tuning: medium wave band with bandspread at higher frequencies (550 to 1,600 KHz).

Earpiece: Magnetic type.

On/off switching: By inserting and withdrawing earpiece plug.

Kit in pack with earpiece, case, instructions and solder £2.48.

Ready built, tested and guaranteed, with earpiece £2.98.

Two Mallory Mercury batteries type RM675 required from radio shops, chemists, etc.

INCLAR MICROMATI

Sinclair Radionics Ltd., London Rd. St. Ives Huntingdonshire PE17 4HJ. Telephone St. Ives (048 06) 4311

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HENRY'S LOW COST FIRST GRADE BRANDED GERMANIUM and SILICON TRANSISTORS, DIODES, RECTIFIERS, DIODES, DI BY ATES - EMIHUS - FAIRCHILD - FERRANTI - I.T.T. - MULLARD - NEWMARKET - PHILIPS - R.C.A. - TEXAS

# TRANSISTORS

A SELECTION FROM OUR LIST

AA.							IK FIZ.	
	¥30 ¥42 Z13	10p	BD115	75p	OC16	50p	2N1303	
44	Y 42	15p 10p	BD123 BD124	85p 80p	OC20 OC22 OC23	85p 50p	2N1304	
III AA	Z15	10p	BD131	75p	OC33	60p	2N1305	
AA	Z17	10p	BD132	80p	OC24	60p	2N 1306 2N 1307	
ACI ACI	26	85p 80p	BDY11	11-50	OC25 OC26	40p 25p	2N1308	05-
AC1	27	25 n	BDY17		OC28	60p	2N 1309	25n
ACI	27 <b>Z</b> 28	50p		11:50	OC29 OC35	60p	2N1613	20p
ACI	28	20p	BDY38 BDY60	65 p	OC35 OC36	50p 60p	2N2147 2N2160	75p
ACI		25p	DD 100	11-00	OC42	40p	2N 2217	25p
AC)	88	2.5 p			OC43	50n	2N2218	20m
ACY ACY	717 718	80p 25p	BF115 BF154	25 p 20 p	OC44 OC45	15p	2N2218	A <sub>OK</sub>
ACY	719	26p	BF158	15p	1 OC70	18p 18p	2N2219	25p 20p
ACY ACY	720	20p	BF159	85p	OC71	19n	2N2219	A
ACY	721	20p 10p	BF180 BF194	85p 15p	OC72	20p 25p	1	25p
ACY	739	55p	BF195	15p	OC76	95-	2N2220 2N2221	25p 20p
■ AC3	740	20p	BF196	15p	0C75 0C76 0C77	40p	2N2221	A
ACS	41	15p	BF197 BFX13	15p	OC81	EUD	ANGOOO	25p
ADI	140	25p 50p	BFX29	25p 25p	OC81D OC81Z	20p 40p	2N2222 2N2222	20p
ADI		50p	BFX30	25 p	OC83	25n	211222	25p
ADI		85p	BFX84	96 n	OC84	25 p	2N2369	15p
ADI		85p 25p	BFX85 BFX86	30p 25p	OC139 OC140		2N2369	A,
AFI	15	25p	BFX87	25p	OC141	40p 60p	2N2646	15p 40p
A F1	16	25 n	BFX88	20n	OC141 OC170	95 n	2N2904	200
AF1	17	20p	BFY18 BFY50	25p 20p	I OC171	80p	2N2904	A
AF1	24	60p 25p	BFY51	20p	OC200 OC201	40p 75p	2N2905	25p 25p
AF1	25	20p	BFY52	20p	OC202	a08	2N2905	A
AF1	26	20p	BFY53	15p	OC206	95p	ı	25p
AF1	27	20p	BFY90	65p 15p	OCP71	97p	2N2906	20p
N AF1	80	30р 50р	B8X20 B8X21	20p	ORP12 ORP60	50p 40p	2N2906	^25p
AF1	81	45p	B8 Y 27	15n	ST140	15p	2N2907	28p
AF1	85	50m	BST95A	12n	ST141	20n	2N2907	A
AF1		40p 40p	B8Y95 BU105 #	12p	TIP29A TIP30A	50p 60p	2N2925	25p 15p
ASY	26	2op	D V 100	15p	TIP31A	60p	2N 2926	10p
ASY	27	2ap 80p	BY126	12p	TIP32A	70p	2N3011	20n
ASY ASY	28	25p	BY124	15p	TIP33A	1 00	2N3053	20p
ASZ	21	80p 55p	BY182 BYZ10	90p 86p	TIP34A	1.00	2N3054 2N3055	50p 75p
BAI	00	15p	BYZII	80p	************************	1.50	2N3525	- 1
BAl	02	80p	BYZ12	avo	T1843	40p		1 00
BA1 BA1	10	25p 7p	BYZ13 BZY78£	25p	T1850 T1851	12p 10p	2N3614	55p
B R A 3	(13	5p	GET102	85p	T1852	10p	2N3702 2N3703	10p 10p
BAX	116	7p	GET103 GET111	25p	TI860	18p	2N3704	12p
BAY	731	7p 15p	GETIII GETII3	45p 25p	T1861 T1862	20p 25p	2N3705 2N3707	10p
BC1	07	10p	GET114	20p	V405A	25p	2N3708	12p 10p
BC1	80	10p	GET115	50p	VR525	35p	2N3714	
BC10	09C	10p 12p	GET116	55p 45p	WO1 WO6	80p 45p	2N 37 15	2.00
BC1 BC1	13	10p	GET880 GET882	30p	ZTX 107	15p		2-20
BC1	14	20p l	GEX66	1	ZTX108	12n	2N3716	
BC1	15 16	20p	GM378A	1.50	ZTX109 ZTX300	15p 12p		2.85
DCI.		900		000				
BC1	16 A	20p 25p	MAT101	ZOD !	LZTX 301	15p		2.00
BC1	16 A	20p 25p 20p	MAT101 MAT121	25p	LZTX 301	15p 20p	2N3772	
BC1	16 A 17 18	20p 25p 20p 20p	MAT101 MAT121 MC724P	25p 60p	ZTX 301 ZTX 302 ZTX 303	15p 20p 20p	2N3772	2·00 2·05
BC1 BC1 BC1	16 A 17 18 19	20p 25p 20p 20p 20p	MAT101 MAT121 MC724P MJ420	25p 60p 80p	ZTX 301 ZTX 302 ZTX 303 ZTX 304	15p 20p 20p 25p	2N3772 2N3773	2.05
BC1 BC1 BC1 BC1:	16 A 17 18 19 34	20p 25p 20p 20p 30p 20p 15p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801	25p 60p <b>80p</b> 80p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501	15p 20p 20p 25p 15p 15p	2N3772 2N3773 2N3791	2 05
BC1 BC1 BC1 BC1: BC1:	16 A 17 18 19 34 35	20p 25p 20p 20p 30p 20p 15p 20p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801	25p 60p 80p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 502	15p 20p 20p 25p 15p 15p 20p	2N3772 2N3773 2N3791	2 05 2 85 2 75
BCI BCI BCI BCI BCI BCI BCI	16 A 17 18 19 34 35 36 37	20p 25p 20p 20p 30p 20p 20p 20p 20p 20p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 £ MJ2901	25p 60p <b>80p</b> 80p 1-10	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 502 ZTX 503 ZTX 504	15p 20p 20p 25p 15p 15p 20p 17p 40p	2N3772 2N3773 2N3791 2N3819 2N3820	2 05 2 85 2 75 35p 55p
BCI BCI BCI BCI BCI BCI BCI	16 A 17 18 19 34 35 36 37 38	20p 25p 20p 20p 30p 20p 20p 20p 20p 20p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 £ MJ2901 £ MJE340	25p 60p 80p 80p 1:10 1:50 50p	ZTX 301 ZTX 302 ZTX 303 ZTX 500 ZTX 500 ZTX 501 ZTX 502 ZTX 503 ZTX 504 ZTX 504	15p 20p 20p 25p 15p 15p 20p 17p 40p 26p	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823	2 05 2 85 2 75 35p 55p 50p
BCI BCI BCI BCI BCI BCI BCI	16 A 17 18 19 34 35 36 37 38	20p 25p 20p 20p 30p 20p 15p 20p 20p 20p 12p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 & MJ2901 & MJE340 MJE370	25p 60p 80p 80p 1.10 1.50 50p 20p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 502 ZTX 503 ZTX 504 ZTX 504 ZTX 504 ZTX 504	15p 20p 20p 25p 15p 15p 17p 40p 26p 7p	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823 2N3824	2 05 2 85 2 75 35p 55p 50p 95p
BCI BCI BCI BCI BCI BCI BCI BCI BCI BCI	16 A 17 18 19 34 35 36 37 38 47 48 49	20p 25p 20p 20p 30p 20p 15p 20p 20p 12p 12p 12p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 & MJ2901 & MJE340 MJE370 MJE371 MJE520	25p 60p 80p 80p 1:10 1:50 50p 20p 80p 75p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 502 ZTX 503 ZTX 504 ZTX 531 IN914 IN916	15p 20p 20p 25p 15p 15p 20p 17p 40p 26p 7p	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823	2 05 2 35 2 75 35p 55p 50p 95p
BCI BCI BCI BCI BCI BCI BCI BCI BCI BCI	16 A 17 18 19 34 35 36 37 38 47 48 49 53	20p 25p 20p 20p 30p 20p 15p 20p 20p 12p 12p 12p 20p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 £ MJ2901 £ MJE340 MJE370 MJE370 MJE520 MJE5221	25p 60p 80p 80p 1 10 1 50 50p 20p 80p 75p 70p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 503 ZTX 504 ZTX 50	15p 20p 20p 25p 15p 15p 20p 17p 40p 26p 7p 10p 7p	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823 2N3824 2N3903 2N3906 2N4058	2 05 2 35 2 75 35p 55p 50p 95p 20p 25p
BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1	16 A 17 18 19 34 35 36 37 38 47 48 49 53	20p 25p 20p 20p 30p 20p 20p 20p 20p 20p 12p 12p 12p 12p 15p	MAT101 MAT121 MC724P MJ420 MJ421 MJ2801 £ MJ2901 £ MJE340 MJE370 MJE371 MJE520 MJE521 MJE521 MJE521	25p 60p 80p 80p 1 10 1 50 50p 20p 80p 75p 70p	ZTX 301 ZTX 302 ZTX 303 ZTX 304 ZTX 500 ZTX 500 ZTX 502 ZTX 503 ZTX 504 ZTX 531 1N914 1N916 1N4148 1844 18921	15p 20p 20p 25p 15p 15p 20p 17p 40p 26p 7p 10p 7p 7p	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823 2N3824 2N3903 2N3906 2N4058 2N4058	2 05 2 35 2 75 35p 55p 50p 95p 20p 25p 12p
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BCI BCI BCI BCI BCI BCI BCI BCI BCI BCI	16A 17 18 19 334 335 337 338 447 448 449 553 554 77 78 9997	20p 25p 20p 20p 20p 15p 20p 12p 12p 12p 12p 12p 15p 20p 12p 15p 20p 15p 20p 10p 10p 10p 10p 10p 10p	MAT101 MAT121 MC724P MJ420 MJ420 MJ2801 & MJ2901 MJE370 MJE370 MJE320 MJE320 MJE320 MJE320 MJE3050 MJE3050 MP8A06 MP8A56	25p 60p 80p 80p 1 10 1 50 50p 20p 80p 75p 75p 75p 25p 25p 25p 25p	ZTX 301 ZTX 303 ZTX 303 ZTX 304 ZTX 500 ZTX 501 ZTX 502 ZTX 503 ZTX 504 ZTX 531 IN916 IN916 IN916 IN916 IN922 2G301 2G302	15p 20p 20p 25p 15p 15p 20p 17p 40p 26p 7p 7p 7p 7p 8p 25p 30p 8p 25p 30p	2N3772 2N3773 2N3791 2N3819 2N3829 2N3824 2N3903 2N3906 2N4061 2N4062 2N4289 2N4289 2N4289 2N4289	2.05 2.35 2.75 35p 55p 50p 95p 25p 12p 12p 12p 12p 12p 12p
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BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1	16 A 17 18 19 334 335 336 337 338 447 449 337 78 79 32L 341.	20p 25p 20p 30p 30p 15p 20p 20p 20p 12p 12p 12p 12p 20p 15p 20p 12p 10p 12p 10p 10p 11p 20p 11p 12p 11p 12p 11p 12p 11p 12p 11p 11	MAT101 MAT121 MC724P MJ420 MJ420 MJ2801 MJ2801 MJE370 MJE370 MJE370 MJE371 MJE521 MJE320 MJE305 MJE305 MP8A06 MP8A56 MP8A56 MP8A56	25p 60p 80p 80p 1 10 1 50 50p 20p 75p 70p 1 10 75p 25p 25p 25p 25p	ZTX 301 ZTX 303 ZTX 303 ZTX 304 ZTX 501 ZTX 502 ZTX 502 ZTX 503 ZTX 503 IN 914 IN 914 IN 914 IN 914 IN 914 IN 912 IN 922 2G 300 2G 303 2G 303 2D 306 2D 306 2D 306 2D 307 2D 307	15p 20p 20p 25p 15p 15p 15p 20p 17p 40p 7p 7p 7p 7p 7p 7p 7p 7p 7p 30p 15p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 11p 20p 20p 20p 20p 20p 20p 20p 20p 20p 20	2N3772 2N3773 2N3791 2N3819 2N3823 2N3823 2N3823 2N3823 2N3903 2N4061 2N4062 2N4062 2N4465 2N4062 2N4557 2N5457 2N5457 2N5456	22.05 22.35 22.75 35p 55p 95p 20p 12p 12p 12p 12p 12p 12p 12p 35p 35p 35p
BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1	16A 17 18 334 335 337 338 47 448 449 553 554 77 78 332L 331L 331L	20p 25p 20p 30p 30p 30p 15p 20p 20p 20p 12p 12p 12p 12p 15p 15p 15p 15p 15p 15p 15p 15p 15p 15	MAT101 MAT121 MC724P MJ420 MJ420 MJ2801 MJ2801 MJE370 MJE370 MJE370 MJE371 MJE521 MJE320 MJE305 MJE305 MP8A06 MP8A56 MP8A56 MP8A56	25p 60p 80p 80p 1 10 1 50 50p 20p 75p 70p 1 10 75p 25p 25p 25p 25p	ZTX 301 ZTX 303 ZTX 303 ZTX 500 ZTX 501 ZTX 501 ZTX 502 ZTX 503 ZTX 503 ZTX 504 ZTX 50	15p 20p 25p 15p 15p 17p 20p 20p 20p 7p 7p 7p 7p 7p 7p 25p 30p 35p 20p 15p 15p	2N3772 2N3773 2N3791 2N3819 2N3823 2N3824 2N3903 2N3903 2N4061 2N4062 2N4062 2N4062 2N4062 2N4063 2N4064 2N4064 2N4064 2N4064 2N4064 2N4064 2N4064 2N4064 2N4064 2N5458 2N5458 40250 40360	2. 05 2. 35 35p 55p 50p 95p 20p 25p 12p 12p 12p 12p 12p 12p 35p 36p 36p 40p
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BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1 BC1	16A 17 18 18 19 334 335 337 338 447 448 449 553 554 57 77 89 89 80 80 80 80 80 80 80 80 80 80 80 80 80	20p 25p 20p 20p 20p 20p 20p 20p 20p 12p 12p 12p 12p 15p 15p 12p 15p 15p 15p 15p 15p 15p 15p 15p 15p 15	MAT101 MAT121 MC724P MJ420 MJ421 MJ2901 MJ2901 MJE340 MJE340 MJE350 MJE350 MJE305 MJE305 MPBA06 MPBA06 MPBA70 MPBA70 MPBU56 MPBU56 MPBU56 MRT214 NKT403 NKT403 NKT403	25p 60p 80p 80p 80p 1 10 1 50 50p 80p 70p 80p 70p 51 1 10 55p 20p 1 10 55p 20p 35p 75p 25p 25p 35p 55p 55p 55p 55p 55p 55p 5	ZTX 301 ZTX 303 ZTX 304 ZTX 504 ZTX 506 ZTX 506 ZTX 506 ZTX 507 ZTX 507 ZTX 508 ZTX 50	15p 20p 20p 25p 15p 15p 15p 17p 40p 26p 7p 7p 7p 8p 25p 30p 15p 15p 15p 15p 15p 15p 15p 10p 12p 10p 10p 12p 15p 15p 15p 15p 15p 15p 15p 15p 15p 15	2N3772 2N3773 2N3791 2N3819 2N3820 2N3823 2N3823 2N3903 2N4058 2N4061 2N4062 2N4289 2N4289 2N4289 2N4289 404030 40361 40361 40362 40430	2.05 2.35 35p 55p 55p 95p 20p 25p 12p 12p 12p 12p 12p 12p 40p 40p 40p 40p
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		8úp	30p		24p	20p
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