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JULY 1971


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## DISPLAYS

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## ADCOLA 64 <br> for Factory Bench Line Assembly

A precision instrument-supplied with standard $3 / 16^{\prime \prime}(4.75 \mathrm{~mm})$ diameter, detachable copper chisel-face bit*.
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WW Iskra high stability carbon film-very low noise-capless con$7.5 \times 2.5 \mathrm{~mm}$. $4 W$ Erie wire wound

| Power | , | wire | Valves |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| wates | Tolerance | Range | available | 1-99 | $100+$ |
| $\frac{1}{2}$ | 5\% | 4.78-2.2MS2 | E24 | $1.0 p$ | $0.8 p$ |
| $\frac{1}{2}$ | 10\% | 3.3M $\Omega-10 M \Omega$ | El2 | 1-0p | $0.8 p$ |
| $\frac{1}{1}$ | 10\% | 18-3.982 | E12 | 1.0p | 0.8 p |
| $\downarrow$ | 5\% | 4-7 $2-1 \mathrm{M} \Omega$ | E12 | $1.0 p$ | 0.8 p |
| 4 | 10\% | $1 \Omega-10 \Omega$ | El2 | $7 \frac{1}{1} \mathrm{p}$ | $7 \frac{1}{2} p$ | order.

DEVELOPMENT PACK
0.5 watt $5 \%$ iskra resistors 5 off each value $4.7 \Omega 2$ to IMS.

E12 pack 325 resistors $\mathbf{£ 2 . 5 0}$.
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$400 \mathrm{~V}: 0.001 \mu \mathrm{~F}, 0.0015 \mu \mathrm{~F}, 0.0022 \mu \mathrm{~F}, 0.0033 \mu \mathrm{~F}, 0.0047 \mu \mathrm{~F}, 2 \frac{1}{2} \mathrm{p} .0 .0068 \mu \mathrm{~F}$, $0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 0.033 \mu \mathrm{~F}, 3 \mathrm{p}, 0.047 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 0.1 / \mathrm{F}, 4 \mathrm{p}$. $0.15 \mu \mathrm{~F}, 6 \mathrm{p} . \quad 0.22 \mu \mathrm{~F}, 7 \frac{1}{2} \mathrm{p} . \quad 0.33 \mu \mathrm{~F}, 11 \mathrm{p} . \quad 0.47 \mu \mathrm{~F}, 13 \mathrm{p}$.
$160 \mathrm{~V}: 0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 0.033 \mu \mathrm{~F}, 0.047 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 3 \mathrm{p}, 0.1 \mu \mathrm{~F}$ $0.15 \mu \mathrm{~F}, 0.22 \mu \mathrm{~F}, 4 \mathrm{p} . \quad 0.33 \mu \mathrm{~F}, 6 \mathrm{p} .0 .47 \mu \mathrm{~F}, 7 \frac{1}{2} \mathrm{p} . \quad 0.68 \mu \mathrm{~F}$, IIp. $1.0 \mu \mathrm{~F}$ 121p.
MULLARD POLYESTER CAPACITORS C280 SERIES
250V P.C. mounting: $0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 3 \mathrm{p} .0 .033 \mu \mathrm{~F}, 0.047 \mu \mathrm{~F}$ $0.068 \mu \mathrm{~F}, ~ 3 \frac{1}{2} p . \quad 0.1 \mu \mathrm{~F}, 4 \mathrm{p} . \quad 0.15 \mu \mathrm{~F}, 0.22 \mu \mathrm{~F} .5 \mathrm{p} . \quad 0.33 \mu \mathrm{~F}, 6 \frac{1}{2} \mathrm{p} . \quad 0.47 \mu \mathrm{~F}$, 81 p. $\quad 0.68 \mu \mathrm{~F}, \mathrm{IIp} . \quad 1.0 \mu \mathrm{~F}, \mathrm{I} 3 \mathrm{p}$.

MYLAR FILM CAPACITORS
$100 \mathrm{~V}: 0.001 \mu \mathrm{~F}, 0.002 \mu \mathrm{~F}, 0.005 \mu \mathrm{~F}, 0.01 / \mu \mathrm{F}, 0.02 \mu \mathrm{~F}, 2 \frac{1}{2} \mathrm{p} .0 .04 / \mathrm{F}, 0.05 \mu \mathrm{~F}$ $0.068 / \angle F, 0.1 / 2 F, 3 \frac{1}{2} p$.
CERAMIC DISC CAPACITORS
100 pF to $10,000 \mathrm{pF}, 2 \mathrm{p}$ each.
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Selection of 100 ceramic and polyester capacitors. 100 pF to $1.0 \mu \mathrm{~F}$, $\mathbf{E 2}-90$ ELECTROLYTIC CAPACITORS-One Price-5p Each
Mullard C426 series $(\mu \mathrm{F} / \mathrm{V}): 25 / 6.4,50 / 6.4,100 / 6.4,200 / 6 \cdot 4,320 / 6 \cdot 4$, $16 / 10,32 / 10,64 / 10,125 / 10,200 / 10,10 / 16,20 / 16,40 / 16,80 / 16,125 / 16$. $6.4 / 25,125 / 25,25 / 25,50 / 25,80 / 25,4 / 40,8 / 40.16 / 40,32 / 40,50 / 40$, $2 \cdot 5 / 64,5 / 64,10 / 64,32 / 64$.
Miniature P.C. mounting ( $/ \mathrm{FF} / \mathrm{V}$ ): $10 / 12,50 / 12,100 / 12.200 / 12.5 / 25$, $10 / 25,25 / 25,100 / 25$.

## POTENTIOMETERS

Carbon track $5 k \Omega$ to IM $\Omega, \log$ or linear (log $\frac{1}{4} W$, lin $\frac{1}{2} W$ )
Single, 12p. Dual gang (stereo), 40p.
SKELETON PRESET POTENTIOMETERS
Linear: $100,250,500 \Omega$ and decades to $5 M \Omega$. Horizontal or vertical P.C mounting ( 0.1 matrix).
Sub-miniature 0.1 watt, 4p each. Miniature 0.25 watt, $5 p$ each.

| SEMICONDUCTORS |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ACI26 15p | BFY52 | 221 ${ }^{\frac{1}{2}} \mathrm{P}$ | OC81 | $15 p$ | 2N3055 | $72 p$ |
| AC127 15p | BSY56 | 30p | OC82 | 15p | 2N3702 | 15p |
| AC128 15p | BS×21 | 25p | ORPI2 | 471 ${ }^{\text {P }}$ | 2N3703 | 14p |
| ADI40 40p | BY124 | $7 \frac{1}{2} \mathrm{p}$ | IN4001 | $7 \frac{1}{2} p$ | 2N3704 | 171 ${ }^{1} \mathrm{p}$ |
| AF115 171p | BYZ10 | 30p | IN4002 | $10 p$ | 2N3705 | 15p |
| AF117 171p | BYZ13 | 20p | 1N4003 | $11 p$ | 2N3706 | $12 p$ |
| BC107 14p | OAS5 | $7 \frac{1}{2} p$ | IN4004 | $12 \frac{1}{2} \mathrm{P}$ | 2N3707 | $18 \frac{1}{2} p$ |
| BC108 10p | OA91 | $7 \frac{1}{2} \mathrm{P}$ | IN4005 | $14 p$ | 2N3708 | 10p |
| BC109 10p | OA 202 | 71p | IN4006 | 15p | 2N3709 | $11 p$ |
| BFY50 22p | OC71 | 15p | IN4007 | $16 p$ | 2N3710 | 12p |
| BFY51 19p | OC72 | 15p | 2N2926 | IIp | 2N3711 | 14p |
| ZENER DIODES $400 \mathrm{~mW} 5 \% 3.3 V$ to $30 \mathrm{~V}, 17 \mathrm{p}$. |  |  |  |  |  |  |
| VEROBOARD 0.15 0.1 |  |  |  |  |  |  |
|  | $0 \cdot 1$ | 0.15 |  |  | 0.15 $521 p$ | 0.1 |
| $2 \frac{1}{21} \times 3 \frac{3}{2}$ | 22p | 16p | $17 \times$ | 31 (plain) | 521 P 37 P |  |
| $2 \frac{1}{3} \times 5$ | 24p | 24p | 17 - | $2 \frac{1}{2}$ (plain) | 3718 |  |
| $31 \times 3 \frac{1}{4}$ | 24P | 24p | $2 \frac{1}{2}$ | 5 (plain) | 1719 ${ }^{\text {P }}$ |  |
| $3 \frac{1}{4} \times 5$ | 27p | 27P | $2 \frac{1}{2} \times$ | $3 \frac{3}{2}$ (plain) | 15 p |  |
| $17 \times 2 \frac{1}{2}$ | 75p | 571. P | Pin in | ertion toal | $147 \frac{1}{2} p$ | 471p |
| $17 \times 3 \frac{3}{4}$ | 100p | 75p | Spot | ce cutter | 3719 | 3719 |
| $17 \times 5$ (plain) | - | 75p | Pkt. 5 | pins | 20p | 20p |

## ROTARY SWITCHES

2P2W, IP $12 \mathrm{~W}, 2 \mathrm{P} 6 \mathrm{~W}, 3 \mathrm{P} 4 \mathrm{~W}, 4 \mathrm{P} 3 \mathrm{~W}, 22 \frac{1}{2} p$
PLUGS AND SOCKETS

| Standard tin screened |  |  |  |
| :---: | :---: | :---: | :---: |
| Standard tin screened | $17 \frac{1}{2} \mathrm{P}$ | 2.5 mm insulated | 1p |
| Standard tin insulated | 14p | 3.5 mm insulated | $7 \frac{1}{2}$ |
| Stereo tin screened | 35p | 3.5 mm screened | 121p |
| Standard tin socket | $15 p$ | 2.5 mm socket | $7 \frac{1}{2} p$ |
| Stereo $\frac{1}{4}$ in socket | $17 \frac{1}{2} p$ | 3.5 mm socket | 71p |

## BRUSHED ALUMINIUM PANELS

$12^{\prime \prime} \times 6^{\prime \prime}=25 p ; 12^{\prime \prime} \times 2 \frac{1}{2}^{\prime \prime}=10 p ; 9^{\prime \prime} \times 2^{\prime \prime}=7 p$.
C.W.O. please. Post and packing, please add 10 p to orders under $\mathrm{C2}$. Data sheets are available for most of the components listed, and will be sent free on request.
8E39 ELSTOW STORAGE DEPOT, KEMPSION HARDWICK, BEDFORD

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12.81 in. Supplied in brand new condition complete fully teatert. 845. Carr. \&1
 meatures ac 97 range instruncolt which Resistance and Power Output liangea d.c. input).
 R.F. measuring heal up to 250 MHz , a.c. watts. 5 watts. Operation $0 / 110 / 200 / 2 \overline{50}$ a.c. cirenit lead and lR.1. probe. \&es. Carr with ADMIRALTY 62B RECEIVERS High quality 10 valve
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Built-in spal phones. Operation $150 / 230 \mathrm{~V}$ a acc. Size $19!$ 13: : 16 in 11 . Weight 11411), Offered \& f . With circuit diagrann. Also avalable B41 L.F. version of above. 1.5 K Hz .700 Hz . 817.50. Carr. E ) j 0 .

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 Illuminated scale. $180{ }^{2} \mathrm{in}$ 230 mm . Weight 81 b .
$220 / 240$ v a.c. Supplied brand uew with hand
book. $\& \& 2$. Carr. 00 p .


## TO-3 PORTABLE OSCILLOSCOPE

 vity $0 \cdot 1 \mathrm{~V}$ p-p/CM. Band. width 1.5 cps-1.jMHz. Input imp. a mega ${ }^{2} \mathrm{zpF}$ X amp. mensitivity 0.9 V $\mathrm{p}-\mathrm{p} / \mathrm{CM}$. Bandwhith $\mathbf{1 - \sigma e p s}$
-800 KHz . Input imp, ${ }^{2}$ -800 KHz . Input imp, ${ }^{2}$ $111 e g ~ a ~ e p p . ~ T i m e ~ b a s e . ~$
$j$ ranges 10 cps- 300 KHz . Jranges 10 eps- 300 KHz . xternal. Illuminated scale $140 \times 215 \times 330$ min. Weight $151 \mathrm{~J}, 2: 2 / 240 \mathrm{~V}$. A.C. Supplied brand new with halubook. 237,50. Carr, J0p.

## $\rightarrow \equiv \mathrm{B}$

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|  |  |  |  |  |  |
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| $1 \mathrm{ma} \quad \cdots \mathrm{l}$ | эА a.c.* | 22.60 | ${ }^{500-0.500 \mu . ~} 81.371$ | T300 dic. | 21.371 |
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Each headphone con tains a 2 lin wooter Built in Individus $25-18,000 \mathrm{c} / \mathrm{s}$. with cable and stereo plug
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1.000 V. a.c. $0 / 50 \mu \mathrm{~A} / 5 / 50 /$ 500 mA . 12 amip. d.c. $0 / 60 \mathrm{~K} / 6 \mathrm{Meg} .60^{\mathrm{o}} \mathrm{Meg}$ n 8.87. Post paid.
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25-80. P. \& P. 15P


TKX MODEL TW-5OK 46 ranges, mirror acale. $50 \mathrm{~K} /$ volts: $0.1+5$ /Volt a.c. D.c. 5. 10. 25, 50, 125, 250, 500 , $1,000 \mathrm{~F}$. A.c. volts: $1 \cdot 5,3$ 5. 10, 25, 50, 125, 250, 500, 1,000V. D. . current: $25,50 \mu \mathrm{~A}, 2.5,5,25,60,250$,
$500 \mathrm{~mA}, 5,10 \mathrm{~A}$. Resigtance: $10 \mathrm{~K}, 100 \mathrm{~K}$, $500 \mathrm{~mA}, 5,10 \mathrm{~A}$. Resistance: $10 \mathrm{~K}, 100 \mathrm{~K}$,
1 meg,
meg. Decibeln:
-20



TE-900 $20,000 \mathrm{n}$ / YOLT GIANT MULTIMETER. Mirrorscale and overload
protection. 6 in full view meter. 2 colour scale. of $3.6 / 10 / 250 / 1,000$ $5,000 \mathrm{~V}$ a.c. $0 / 25 / 12 \cdot 6 / 10 / 50 / 250 / 1,000$ , 000 V d.c. $0 / 50 \mu \mathrm{~A} / 110 / 100 / 600 \mathrm{~mA} / 10 \mathrm{~A}$ d.c. $02 \mathrm{~K} / 200 \mathrm{~K} / 20$ nues. ohm. 215.
P. $\&$ P. 25 p .

$\begin{array}{lr}\text { MODEL } & \text { 6025. } \\ \text { ranges, } & 87 \\ \text { giant } & 5+1 \text { in }^{2}\end{array}$ $\rightarrow \begin{aligned} & \text { ranges, giant stin } \\ & \text { meter, polarity }\end{aligned}$ reverse switch. Sen sitivity: $50 \mathrm{~K} / \mathrm{Vol}$ D.c. Tolts: $0.125,0.25,1.25,5,10,25,50$ 0,25, , 100, . 1 . Volte: $1-5,3,5$ $10,25,50,125,250,500,1,000 \mathrm{~V}$. D.c current: $25, ~$
$500 \mu \mathrm{~mA}, 2 \cdot 5,5,25,50,200$
$5,10 \mathrm{~A}$. Reaistance: $2 \mathrm{~K}, 10 \mathrm{~K}$ $100 \mathrm{~K}, 1$ meg, 10 meg. Decibels: - 20 K



MODEL TE12. 20,000 $\begin{array}{ll}\text { O.P.V. } \quad 0 / 0 \cdot 6 / 30 / 120 / 600 \\ 1.200 & 3.000 / 6.000 \mathrm{~V}\end{array}$ $1 / 6 / 30 / 120 / 600 / 1,200 \mathrm{~V}$ a.c / $60 \mu \mathrm{AA} / 6 / 60 / 600 \mathrm{MA}$
$0 / 6 \mathrm{~K} / 600 \mathrm{~K} / 6 \mathrm{meg} / 60$. Megohns 50 PF . $6 \mathrm{meg} / 60$ 26.071. P. \& P. 17tp.

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$10-0,100 \mathrm{mfd} .0-100-0 \cdot 1 \mathrm{mfd} \cdot 23.47 \mathrm{t}$ P. \& P. 15 p .


MODEL TE-90. 50,000 oad protection. 003/13/60 $00 / 600 / 1,200 \mathrm{~V}$ d.e $0 / 6$ $30 / 120 / 300 / 1,200 \mathrm{v}$ 0/6/ 03/6/60/600m1 $6 \mathrm{~K} / 160 \mathrm{~K} / 1 \cdot 6 / 16$ meg. -20 to +63 dB 7.50. P, \& P. 15p.

TME MODEL TW-20CB. eatures Resettable Over
fad Button. Sensitivity $20 \mathrm{~K} \cap /$ Nolt d.c. $\$ \mathrm{KQ} /$ Yolt a.c. I.c. volte: $0-0.5$. -5, 10, 50, $260,1,000 \mathrm{~V}$ : A.c, volts: $0-2 \cdot 5.10,50,250,1,000 \mathrm{~s}$. D.c. $\begin{array}{llllll}\text { currents: } & 0-0.05, & 0.5 & 5, & 50, & 500 \mathrm{~mA} \\ \text { Resistance: } & 0-5 \mathrm{~K} & 50 \mathrm{~K} & 0.500 \mathrm{~K} & 5\end{array}$ Resiatance: $0-5 \mathrm{~K}, 50 \mathrm{~K}, 0-500 \mathrm{~K}, 5$ meg 171p: -20 to +52 dB . 21150

MODEL A8-100D. $100 \mathrm{~K} \Omega$ Volt. Sin, mirror 8cale. Built-in meter protection 0/ 3/12/60/120/300/600/1,200V d.c. $0 / 6 / 30 / 120 / 300 / 600 \mathrm{~V}^{\circ}$ $\begin{array}{ll}\text { a.c. } & 0 / 10 \mu \mathrm{~A} / 6 / 60 / 300 \mathrm{nA} / \\ 12 \mathrm{~A} . & 0 / 2 \mathrm{~K} / 200 \mathrm{~K} / 2 \mathrm{M} / 200 \mathrm{M}\end{array}$
 +17dB. 21280. 100,000 fin scale buzzer hort circuit check. Sensitivity : 100,000 OPV'd.c. $\delta /$ Folt a.c. $0,50,250,1,000 \mathrm{y}$
 A.c. volts: $3,10,50,250,500,1,000 \mathrm{~V}$ D.c. current: $10,100 \mu \mathrm{~A}, 10,300,500 \mathrm{~mA}$ $2 \cdot 5,10 \mathrm{~A}$. Reajstance: $1 \mathrm{~K}, 10 \mathrm{~K}, 100 \mathrm{~K}$ 10 meg. 100 mer. Decibela: - 10 to




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Bands covering $550 \mathrm{KHz}-30 \mathrm{MHz}$. B.F.O Buit in Speaker $220 / 240 \mathrm{~V}$ a.c. Brand new with inatructions. 215.76. Carr. 37 ip .

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## THAT SPECIAL PROBLEM

I is now becoming widely appreciated that many private constructors are highly competent, expect up-to-the minute designs, and that they generate a demand for many new and unusual ty'pes of components, such as industry had previously deemed its exclusive own. This is the heartening news to emerge from correspondence arising from our comments in the May issue on the subject of component availability. On the less bright side, problems of supply there are, and will remain; but all concerned-makers of components, industrial distributors, and retailers-seem keen to mollify them.
Several large component manufacturers have made clear that no embargo is imposed by themselves that would prevent their products ultimately reaching the hands of private individuals. The operative word here of course is "ultimately". In the majority of cases the components must pass through the established distribution system. The terminal point so far as the private purchaser is concerned is the retailer or, exceptionally, an industrial distributor.
A significant development, in fact, is the interest certain industrial distributors are showing in the growing amateur market. Some are already entering the retail mail order business. Is this a challenge to the established component retailer? Some retailers admit to us that the supply of "specials" is an uneconomical business, but they are prepared to do it as a service to their customers.
A few retailers, resenting what they felt to be an attack upon their business acumen in our May comment, have laid the blame for much of the present difficulty upon Practical Electronics. Two main points have been raised. They are of interest to anll involved in amateur electronics, and we will give our answers to both points herewith.
A common (and perfectly understandable) request, is for advance notification of components that will be specified in forthcoming articles. Admirable as this sounds, it overlooks the exigent demands that arise in producing a monthly magazine containing extensive technical detail. Could such problems be overcome, there would remain the severe burden of circulating all advertisers each month: it would be an invidious act to select just a few.

A second criticism levelled at this magazine is that we should make certain that all items specified are actually available to private purchasers. Suppliers and constructors alike can be assured that great pains are taken to verify this. Often alternatives are substituted prior to publication when perusal of retailers' current catalogues

## THIS MONTH

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[^0]THE simplest form of intercommunication system is made up of one master unit and one slave unit. The slave unit consists of a loudspeaker serving in the dual role of microphone and loudspeaker.

For the other unit the choice of title "master" derives from the fact that all two way conversation is controlled from it as seen in Fig. 1. Here, master loudspeaker LSI is connected by a short twin lead to the amplifier via a two way switch. In the "Talk" position the switch connects LSI to the amplifier input and the remote slave loudspeaker to the amplifier output.

It is common practice to have the function switch biased to "Talk" as all conversation is initiated by the master. However, as the intercom might find use as a baby alarm or doorphone this convention was dropped.

As a microphone, LSI acts as a voltage generator with the voice coil responding to sound waves striking the cone. The amplifier steps these voice signals up and relays them to the slave load LS2. A person


Fig. I (a). In the "Talk" position master loudspeaker LSi acts as a microphone. LS2 functions normally as a loudspeaker (b). In the "Listen" position, slave loudspeaker now functions as the microphone and initiates the call. Ideally screened lead should be used for LS2 to eliminate noise influences


Fig. 2. Circuit diagram of intercom unit
in proximity to this loudspeaker responds by simply speaking up when the master unit switch is set to the "Listen" position.

## COUNTING THE COST

Probably the most important criteria for an intercom system is low running and building costs. Whilst a rectified mains supply will effectively reduce running costs by avoiding batteries, it does introduce the hazard of possible accident, the likelihood of hum on the line and the necessary siting of the master unit near a mains outlet.
Anticipating the manifold applications of the system in garden workshops or garages, battery power was considered a prime requirement. In fact quiescent current drain is about 8 mA with peaks, according to speech level, of about 35 mA .
The cost of the unit to build should be around $£ 1 \cdot 50$. This excludes the price of the loudspeakers.
Most intercom systems provide peak power outputs from about 100 to 500 milliwatts. To the hi-fi man in pursuit of an ever spiralling power figure this must represent an almost inaudible sound level when coming through a loudspeaker. In fact, a power output of 100 milliwatts is adequate for room listening.

Equally, whilst a sound purist might reach for his hat when you beg to demonstrate your intercom amplifier with a harmonic distortion content approaching 10 per cent, impress on him that an intercom is for intercommunication and the intelligence loses little by this type of distortion.

## CIRCUIT ACTION

In Fig. 2 is given the intercom circuit diagram with T-Dec hole positions for plugging in component parts. If translation of this unit into the form of a more permanent assembly is intended, such as on Veroboard, then the prototype component geometry should be maintained to lessen the likelihood of instability.

The circuit diagram shows S1 in the "Talk" position so that the call facility is with the master unit. Cl provides d.c. isolation so that no biasing of the loudspeaker cone is possible.

Since the loudspeaker in use is a voltage generator of low output impedance it must work into a higher load to effectively transfer small signals. TR1 is a silicon transistor which means that the simplest form of base biasing can be used as the expected small changes in surrounding temperatures are hardly likely to affect its working point.

## Components

| Resistors |  |
| :--- | :--- |
| R1 | $100 \mathrm{k} \Omega$ |
| R2 | $2.2 \mathrm{k} \Omega$ |
| R3 | $22 \mathrm{k} \Omega$ |
| R4 | $2.2 \mathrm{k} \Omega$ |
| R5 | $270 \Omega$ |
| R6 | $100 \mathrm{k} \Omega$ |
| All | $10 \% \frac{1}{2}$ watt carbon |


| Capacitors |  |  |  |
| :--- | :--- | :--- | :---: |
| C1 | $4 \mu \mathrm{~F}$ | elect. I5V |  |
| C 2 | $4 \mu \mathrm{~F}$ | elect. I5V |  |
| C 3 | $0.01 \mu \mathrm{~F}$ | polyester |  |
| $\mathrm{C4}$ | $0.01 \mu \mathrm{~F}$ | polyester |  |
| C5 | $250 \mu \mathrm{~F}$ | elect. 15 V |  |
| C6 | $100 \mu \mathrm{~F}$ | elect. I5V |  |
|  |  |  |  |
| Transistors |  |  |  |
| TR1 | ZTX 300 |  |  |
| TR2 | 2N2926 green spot |  |  |
| TR3 | ZTX 300 |  |  |
| TR4 | ZTX 500 |  |  |

## Switches <br> S1 D.P.D.T. roggle <br> S2 On/off toggle

Loudspeakers LSI, LS2 $5 \mathrm{in}, 15 \Omega$
(2 off)

## Miscellaneous

BY|-9V
T-Dec, connecting wire


Fig. 3 (a). Transistor output characteristic showing two common bias conditions (b) How input characteristic non-linearity produces crossover distortion with zero biased input signal (c) sllght forward bias for the complementary transistors gives a clean output voltage ( $V_{\text {out }}$ )

Both this and the succeeding transistor TR2 work in a condition known as Class A. In Fig. 3(a) is shown an output characteristic for a transistor with a resistance load. If a signal current is passed to the base this will be reproduced at the collector ( $V_{\text {ont }}$ ) only if it is well contained within the load line. For small signals, Class A, or mid point bias, does in fact provide this as can be seen.

Since we are not concerned with power in the first two transistor stages the inefficiency represented by Class A biasing can be ignored. However, where transistors are used as power amplifiers. with large collector current swings, mid-point biasing would mean a large continuous drain from the power source and the batteries would soon be flat.

## CLASS AB

Biasing further down the load line to near cut-off will give a very low quiescent current drain but intolerable distortion as the collector voltage is only reproducing half the signal voltage. Biasing at this point is known as Class $A B$.

If an npn/pnp combination are arranged as TR3/TR4 in the circuit and the gain characteristics of these transistors matched then we have the advantage of excellent efficiency and low distortion in a power output pair. As previously stated, since second harmonic distortion in speech is not a cause for concern the cost or problems of matching, can be ignored.

## CROSSOVER DISTORTION

There is a form of distortion found with this type of output configuration which is intolerable. Known as crossover distortion, it occurs when the transistors are zero biased as seen in Fig. 3(b).

Here a sinusoidal signal applied to the transistor bases will be distorted at the output due to curvature on the input characteristics. At low signal levels this distortion is particularly bad.

To overcome this a small amount of bias is applied to both transistors so that any signal transfer is made on the linear part of the input characteristics. Fig. 3(c) shows how effective this is when related to the output characteristics.

In the circuit diagram VR1 is used to make this bias adjustment.

## NOISE

In any high gain amplifier there is always the possibility of noise being introduced, this applies particularly where long runs of wire are introduced directly to the preamplifier input. With an input impedance of less than 1 kilohm the noise problem, in general, should only be a small one even for substantial twin runs to the slave loudspeaker.

It must be realised, however, that noise will raise the quiescent current level. Since this is normally about 8 mA single screened lead connected as in Fig. 1 will prove effective against such influences.

Spontaneous high frequency oscillation can also be a nuisance so decoupling capacitors C3 and C5 are included.

## SETTING UP

Since there is only one control requiring preliminary adjustment, namely VRI, setting the amplifier up is simple and only requires a multimeter. First set VR1 wiper to mid travel. With S1 in the "Talk" position connect the multimeter in series with the battery with range switches set to 100 mA d.c.

With the unit switched on the current reading should be about $20-30 \mathrm{~mA}$. Now adjust VRI to the lowest current reading which should be about 8 mA . If now a transistor radio is placed before LSI, preferably tuned to some speech programme, VRI should be adjusted for minimum distortion at the slave output LS2. If the radio volume is set too high you will probably overload the amplifier as it has an input sensitivity of about 1 mV r.m.s. to provide an output of 120 mW .

Turn the radio off and check the quiescent current, this should not have moved much from the 8 mA figure. With these procedures satisfactorily carried out, the intercom should be ready for use.



The command module pilot, astronaut Roosa, took thousands of photographs of the moon's surface and one of these, on the far side of the moon, showed a new crater which is probably the youngest on the moon.

On the near side, another new feature was detected near the Crater Lansburg which looks like a winding ditch. It is not a rille or fault line and the astronaut has called this feature "The Thing".

## CANADIAN SATELLITE

Another Canadian satellite of the ISIS series launched by America triggers off another international scientific project. Code named ISIS 2 it weighs 260 kilograms and will, for the first time in the series, send continuous "ionograms" for display on television consoles.

The name "ionogram" is given to the picture which records the depth and density of the ionosphere. Sudden changes in density or depth as viewed from above can give direct indication of the propagation conditions for radio communication.

The first $I S I S$ satellite launched in 1969 is still in operation but only transmits ionograms at specific periods, most of its data is in the form which requires processing at ground stations in order to arrive at a picture.

Previously the ground station made an ionospheric sounding as the satellite passed overhead and two sets of data were correlated and a picture built up. This results in a delay in warning when ionosphere disturbances arise.

## TRACKING CENTRE

The ISIS 2 satellite is of a more advanced design containing a new telemetry system where the ionograms are relayed directly to each participating centre, together with other data, with the result that immediate knowledge is available of the ionospheric condition.
The stations involved are the tracking stations in America, Canada, United Kingdom, Norway, Japan, India, Australia and New Zealand. The information is also deposited at the World Data Centre at Boulder, Colorado.

The advanced ISIS 2 is the forerunner of a series of ionospheric operational satellites and these together with a world wide network of ground stations will be of the utmost value to communications operations, enabling a minute by minute decision to be made as to the choice of the best frequency for propagation.
Other data provided by these satellites will include information of
the effect of the sun on the earth's atmosphere and may lead to the solution of how atmospheric pollutants are disposed, and how and what mechanism might be available naturally for dispersal.

## LUNAR NAVIGATION

The two astronauts Shepard and Mitchell in their report on the Apollo 14 mission have made the point that personal navigation on the Moon needs instrumentation.

Part of their work task was to reach the Cone Crater in the Fra Mauro Region, but they turned back before reaching it as they thought there was not sufficient time to carry out this operation. In actual fact they were within 50 feet of the crater but the difficulty of deciding on distance made it impossible for them to know this. The horizon is so close, there are no colour differences to compare distances against. and the lack of contrast on the surface indicates that some kind of range finder is required.

## LUNAR LOPE

The astronauts said that the most difficult thing they found in their walks was this inability to judge distance accurately. They thought that they were covering more distance than in fact they were because of this, and also that in doing the "Lunar Lope" of one step and a hop they did not move more than three or four feet a second; which is only slightly faster than walking normally on Earth. This bothered them because they felt in the one sixth gravity condition on the Moon that they were moving much faster than this.

The long distances that will be covered by the crews of Apollo 15 and 16 using the "lunar rovers" will be accurately indicated on the vehicle instruments.

The report also noted that the magnetometer which they set up on the Apollo 14 mission has indicated that the magnetic field in the Fra Mauro area is much higher than that found at the other landing sites visited by $A$ pollo 11 and 12 missions.

## ROCK COLLECTION

It seems that some of the rocks collected from the moon are at least 4,500 millior years old and must belong to the original crust. This was the preliminary statement made by Paul Gast and Robin Brett of the Manned Spaceflight Centre at Houston.

The samples brought back differ considerably from those of the Apollo 11 and 12 missions, and also from the Luna 16 mission of the Soviet Union. For one thing they are more complex, containing more minerals and so far 23 different minerals have been detected, though ten of these have not yet been positively identified. It is also apparent that these rocks reveal several different phases of the moon's
history. history.

Another difference between the Apollo 14 and previous missions for rock collection is that the latest samples contain very few fragments of igneous rock but they do contain some of the elements that formed the Earth's crust.

It is possible to learn more about the earth from the moon samples since they are probably in exactly the same condition as when they were formed, whereas on earth much of the geological record has been obliterated.

## MOON BIRTH

A theory has been advanced that the moon was formed some 4,500 million to 5,000 million years ago by planetoids composed of the lighter gases that were thrown off from the coalescing, earth and formed a ring round it. It is also suggested that another of the planetoids moved in an erratic path and when the earth and moon became distinct and separate it crashed on to the moon to form the oldest maria the Mare Imbrium or Sea of Rains.
Whether this is a tenable theory or not may emerge from the different centres to whom parts of the moon rocks have been distributed. There will be more than 700 investigators in some 187 groups all over the world working on these samples.

This is one aspect of the value of the manned missions to the moon and later other planets.

## MATRIX DISPLAYS

In the preliminary design work on the "P.E. Aurora" unit a simple control system with a matrix of lamps was used as the basis for the switching control. The first arrangement made is shown in Fig. I, where 16 bulbs are set out on a $4 \times 4$ grid. The control system described previously in Part 1 was such that, in order to light up the bulbs, it was necessary to power them through both co-ordinate axes ( $x$ and $y$ ).

Experiments showed that there is a great potential in programming lights by the use of a matrix. Besides the obvious.varations in thyristor controlled a.c. the disposition of the matrix itself is an important visual aspect of the whole system.

If the point sources in Fig. 1 are changed to small tungsten strip lights, the pattern configurations are radically altered.

The grid may be disposed in many ways Fig. 2 shows, for example, an arrangement where one coordinate is radial and ane concentric.

Fig. 3 combines three co-ordinates, two being curved radials, and the third eccentric. Visual effects achieved could be rotation and counter-rotation or expansion and contraction from the centre.

A light display may be three dimensional in form, and the matrix applied as in Fig. 4.

## ,



Fig. 1. Basic matriz of lt lamps triggered from an eight-chanmel controller PART 4-LICHT DISPLAYS BY M.LEONARD A.RIB.A

## BULB SELECTION

Examination of manufacturers catalogucs bill show bulbs of various shape and size. It is suggested that the reader obtains a number of hulbst in differing forms for experiment. It is advisable to obtain? prices before committing oneself to a particular arrangement, as some bulbs can be very expensive when buying eight or even 16 of them.

The characteristics of a particular bulb will "often be the starting point in the design of a dipslay. Many bulbs are too bright to be left exposed, but those of lower wattage or of opal finish may be glare free.

Bulbs like quartz-iodine have small point source and throw sharp shadows, whereas strip-lights obviously give a more diffused light. Silver spot bulbs are internally mirrored at the base end in order to throw light forward as a beam; whereas "crown silvered" are mirrored on the top. and thus send light back to the socket end of the bulb at which there is normally a reflector. This can be adjusted to vary the focus of the beam.

In the course of selecting suitable bulbs, take heed of the restrictions imposed by the power ratings of the thyristor controller tise Part 11. Examination of some of the better designed light fittings in worthwhile, for these show how a great variby of control can be achieved over a source as mundane as the domestic light bulb. Certain fittings may give a soft diffused light, others may throw is texture of light and shade on to surrounding surfaces.
Glass elements may be incorporated to simulate the glistening chandelier, or a bright light screened by metal in which there are a number of fine perforations - this giving a brilliant sparkle of light.

Fig. 3 (right). Three-dimensional display taking in four concentric channels, 12 A radial channels, and 12 B radial channels (total 48 lamps, $2 \cdot 16$ control channels)


Fig. 4. The lamps can be re-arranged to suit geometric outlines such cs a cone or sphere with latitude and longitude control channels


## DISPLAY CABINET

In the final display cabinet, as illustrated on the front cover and in Fig. 6, the bulbs used were 40 watt 240 volt "continental" fittings, of $1 \frac{3}{4}$ in diameter. These were "pearl" in finish and were left exposed, creating perfect spheres of white light.

Fig. 5 shows the plan arrangement of a single bulb, set within a fan of coloured transparent plastics sheet. The surfaces, being highly reflective, act as mirrors; the image of the bulb appears at points A, B, C, D. Each reflected image appears to be a different colour, and the colours will be different from different viewpoints.

At the rear of the display was a glass mirror so that the form which was a half cylinder appeared reflected as a complete cylinder. Lights arranged in spiral form (Fig. 6) gain the best effect from the multiple reflections, only half the spiral being constructed.

Wiring was taken from each lamp holder. via a stalk to the main vertical support column and then to the backboard of the structure. The stalks consisted of metal tubes fitted at one with the lamp holders, and the other end fixed to a common large diameter central trunk, through which all the supply lines are fed. The ballast lamps shown in the matrix in Part 1 can be incorporated as permanent lamps shown in the display.


Fig. 5. One bulb can supply a large number of images by reflections in Perspex panels in the shape of a fan

## COLOUR MIXING

A very important principle in the control of colour is the use of additive or subtractive colour mixing. In mixing paint the primary colours are red, blue and yellow. The colour seen from a painted surface is that component of white light left after the pigment has absorbed the remaining part of the spectrum. Combinations of pigment may be made which progressively absorb more and more of a white light source, until black is reached; at this point all light is absorbed and none reflected.

The effect achieved by overlaying transparent colour filters is also a subtractive one. Thus the mixture of blue and yellow paint, or the overlay of blue and yellow filters will produce green. It is important to remember this principle to distinguish colour light from colour pigment.


Fig. 6. The finished display cabinet was bullt around a central trunk with radiating fins of coloured Perspex and a mirror at the back. Matrix wiring is used as in Part 1 , the diodes being molinted on the back panel


Fig. 7. Tri-colour shodow casting technique

## COLOUR LIGHT MIXING

However, when coloured beams of light are overlayed the effect is completely different, for the principle is that of additive mixing. The light primaries are red, blue and green. Those readers who try this technique for the first time may be surprised to find that, for example, the addition of red, and green light produces yellow light.

If three light sources which are the light primaries are arranged to illuminate an opal screen and completely overlap one another (this could be a flat surface, a sphere or cylinder) and each bulb controlled in light intensity, variations in intensity will produce a complete spectrum of colour. In the first instance controls could be operated manually, and later, sequences of colour change programmed electronically.

Fig. 7 shows another way in which the three light primaries can be used. An object A is lit by a red, blue and green light. The red light will illuminate the wall behind the object, casting a black shadow.

The addition of the green light will overlight the shadow turning it green, and where red and green light mix, the product will be yellow. The addition of blue will increase the complexity of colours and shadows. Movement of the lights will cause further complications if the shadows are allowed to overlap.
This experiment is a very rewarding one and can be translated in a number of ways. If the object is a dancing figure, and the light primaries provided by theatre spot lights, dramatic effects of colours and shadows can be back projected onto a translucent screen. Alternatively, the whole idea can be miniaturised in the form of a small light box.


Fig. 8. Colour reflecting from the idea shown in Fig. 7. The coloured light is projected onto a translucent screen of opal finish

## LIGHT MURALS

A number of methods have been used to make light murals; the main problem is how to keep the depth behind a screen surface to a minimum. One system shown in Fig. 8 is to replace the object illuminated with a series of rotating shaped reflectors.

Light sources are beamed in from the edge of the screen, and coloured reflections are thrown forward; mixing taking place on the screen can be reduced to a few inches. One example of a system similar in principle to this is the "Dreamscreen" (see Electronorama last month).

Fig. 9 outlines a completely different approach. The elements are arranged in layers. The rear layer is that of the light sources, which can be white, coloured, or combinations of both. In front of this are a series of cut out shapes which can be opaque, or translucent colour filters, or again combinations of both. The next layer can be hardboard, which is obtainable with a variety of perforations; reeded or cross reeded glass gives further interesting effects.


Fig. 9. Back projection onto an opal finish screen through layers of translucent material or perforated sheet

The final layer is an opal screen. A simple framework will be needed to allow the planes to be suspended, and the distance between the elements easily varied.

Switching lights in sequence will give a movement of shadows and colours on the front screen. The perforated layer can give a pinhole camera effect focusing images, the definition of which will depend on size and shape of the holes. In some instances this layer may work best close to the bulbs. The reeded glass can act like a series of lenses, and often the opal screen can be omitted, and the effects of the programmed lighting seen directly in the glass layer.
The light beam of a slide projector will give a whole new range of effects if used to replace the light sources shown in Fig. 9. Optical effects can be introduced in front of and behind the lens.

## PLAN A SYSTEM

The foregoing ideas are based mainly on geometric arrangements, and will in some way produce a formal display restricted only by the boundary of the
*


Fig. 10. Simple wall of wrinkled aluminium foll for direct front projection of coloured light
arrangement. An example of their applications would be as an exhibition centre piece. Several other arrangements can be devised and it is well worth the constructor's time to sit down with pencil and paper and plan a system suitable to the setting.

If a less formal display is required, maximum use can be made of walls, ceilings or other fixtures to convey the light patterns.

## ALUMINIUM WALL OR SCREEN

One experimental system that was tried successfully at Sound '7l exhibition was a wall of wrinkled aluminium foil (Fig. 10). A simple timber framework was used to hang domestic cooking foil; heavy duty foil is recommended and in fact a caterer's pack size would be particularly suitable.

The foil is hung from the top cross bars of the frame, fixing with transparent adhesive tape. Try not to fold over or crease the foil during the setting up, otherwise undesirable crease lines will remain.

When hung, just tack together the adjacent vertical edges of the strips to hold the whole screen fairly stable. The strips of foil could be made to extend about four to six feet along the floor in front of the foot of the wall. This will then give
the appearance of an extended display showing reflections of light in the floor.
If possible, the wall of foil should be made slightly concave to help spread the light projected onto it. When set up, the foil is wrinkled and dented by gently tapping all over with the flat palm of the hand. The larger the indentations are, the better will be the final effect.

## EIGHT CHANNEL OPERATION

The lights are now set up for straight eight-channel operation (not matrix), i.e. one bulb per channel.
Although expensive to purchase initially, it is well worth obtaining colour flood lamps (e.g. Philips "Comptalux" E27) rated at 240 V 100 W a.c. These can be obtained in red, blue, green and yellow and clear, with reflective rear inside surfaces and diffuser lenses, and will provide the basis for some interesting colour mixing as described earlier. These lamps use Edison screw bases, which are mounted on either side of the foil screen, either on wooden battens or in cases.

The spacing of the lamps will best be found by experiment. They can be mounted close together about six feet away from the foil, but the light will be restricted to a small area on the foil (Fig. 11). However, by doing this, some colour mixing can be achieved.

If there is an apparent gap in the centre of the screen, use two lamps to project from the centre near the floor. The small pictures on the front cover will give some idea of what can be done.

## ADDING VARIANTS

There are some simple variations that the reader will want to try, such as suspending twists of foil strips, each about 2 in wide from cotton just in front of the screen. Any ambient heat in the room will encourage the twist strips to rotate while the light shines on them.

The screen could be made entirely of freely suspended narrow strips that can move, but no movement should be violent or the effects will be spoilt.


Fig. II. Suggested positions of the eight coloured lamps in relation to the aluminium wall


Fig. 12. Staggered foil strips on a bin wide supporting batten


Fig. 13. 'Cinemascope" concave mirror of foil

Try staggering the positions of the large foil strips on either side of a 6 in board (Fig. 12).

One of the simplest and cheapest ways of displaying the light is by projecting across a ceiling, especially one that has the modern stipple finish texture. The lamps are fitted below ceiling level, perhaps on the top of a pelmet. The result is a changing ambient coloured light that is not too violent to the eyes, but spread over large areas of the ceiling.

## SUBTLE CHANGES

When setting up any colour light display, always bear in mind that the "P.E. Aurora" system was designed for subtle changes in light intensity without flashing. The degree of change and the speed of switching will depend on the music driving the system, but overall one should not expect to get harsh flashing unless the preset controls in the electronics are incorrectly set up.

Follow the setting up instructions given in the earlier articles, and experiment fully with light positioning. Earlier notes on using patterned glass or coloured Perspex apply equally well, whatever the type of display you choose.

In any house there are many objects, from cheesegraters, to glass ash trays which can be used as a starting point in the experiments of displaying light. From these simple beginnings the reader can progress to the manufacture of optical elements more tailored to his own needs and finally to systems of switching and dimming of the light sources.

Next month's article will discuss some alternative methods of driving the lamp controller and included will be a random digital sequencer which triggers the lamp channels in all possible sequence codes.

The best display for this method, which incidentally does not require the use of filter circuits, is to matrix the channels to get maximum number of bulbs operating. This is typically a "Christmas Tree" effect suited to displays of random flashing lights, and is ideal for the geometric displays described earlier in this article.

Note: In Fig. 21 last month (page 505), C19 should be C24.


Fig. 14. Convex side projection


Fig. 15. Rotating reflective shapes in front of a curved white screen

## Gerry Brown Wifferwil BITTER SWEET

When next you are mucking about in your workshop, or study, with an audio frequency tone generator do be careful to make certain that you don't happen to be eating, or swallowing a mouthful of beer, at the same moment. Why? Well you just might be on the wrong frequency!

Sounds idiotic. I know, but the Danish psychologist, Dr Kristian Holt-Hansen, has recently indicated that there definitely seems to be a strong connection between the perception of sounds and taste.

In his experiments, subjects have reported that quite widely varying foods and drink seem to relate with specific frequencies at which they taste best. Apparently, when the pitch of the sound was taken either above or below this specific tone the taste became less pleasing; furthermore, a unique "harmonytone" could be shown to exist for each taste.

The really interesting point, I think, is that the subjects were almost unanimous in agreeing that each food had a tone-pattern which was the same for every person being tested.

With the diminishing prices of i.c.'s, one can conjecture that it will not be long before some enterprising individual puts a vinegar on the market that has all the bouquet of a good wine provided that the battery, driving an oscillator built into the bottom of the bottle. doesn't go flat!

## CHICKEN

Don't ask me why, but a farmer friend of mine was telling me, over a glass of beer the other day, that he wanted to be able to determine the movements of a particular cockerel he owned.


However, this had caused him some head-scratching at the time, because a photo-cell technique would have been useless in the application due to its inability to discriminate between one bird and another. He had also thought of using a radio-active tracer, but decided against that on the grounds of contamination.

His final solution to the problem was to attach a tiny magnet a little way up the chicken's leg and use this to cause a movement on the needle of a compass buried a couple of inches in the ground. The needle movement was sensed by a simple pick-up coil and the output fed to a low-frequency amplifier connected to a delay circuit and relay, Fig. 2.

Whenever the cockered walked over the compass the needle would spin and thereby induce a current in the sensing coil. After amplification and a short time delay it operates the relay which activates a counter.

I understand that this fowl plan was eminently successful!


## PSI EFFECT

One can never fail to admire the fortitude of the psychic researcher. An example which perhaps typilies what I mean has been set recently by Professor Rhine of Duke University, North Carolina.

Because the rules of human studies in the ESP (extra sensory perception) field have proved so indeterminate, he has chosen to use mice in his studies.

In the tests, the mice are placed in a cage which has a low lengthwise division which forms a barrier the mice can jump over easily when cicaping from a harmless electric shock applied randomly through the floor on one or other sides of the partition. Although the shocks are applied randomly, relative to sides, they occur at regular times and so the mice have the chance to develop "hunches" about which side of the barrier the shock will occur and either jump or not, as the case may be.

Statistically they ought to be correct 50 per cent of the time, but (mice being what they are) this worked out to be 54 per cent! Precognition? Rhine seems to think so; in fact he goes so far as to say that he believes the whole of the animal world may have this power.

Who can say; certainly it's beginning to look as though there might just be something in what he says.
Believe it or not. I was reading a paper only today, by J. L. Randell, entitled "Experiments to Detect a Psi Effect with Small Animals". And the animals? Would you believe woodlice?

Suffice to say that these experiments appear to have proved positive and are backed by good solid mathematical logic.


## ROUND THE BEND

Motorcyclists have a term for a particular form of enjoyment known to their fraternity as "bendswinging". Simply, this amounts to laying one's machine over at an angle approaching the horizontal when navigating the various bends in the road.

Whilst this kind of amusement can be dangerous, even for motorcyclists, the similar practice of "taking "em as fast as you can" could be positively prohibitive in a car when one has little or no idea of the banking-angle/speed required to turn the vehicle completely over!

Since it is difficult to determine just how close a car is to instability at such times the little device given in Fig. I should help provide the warning required. The device comprises a pair of mercury switches, suitably mounted on the bulkhead of the car, interconnected with some form of "attention-getter" like a buzzer or an oscillator.

In use, if one enters a bend too fast without much conscious thought, the centrifugal force acting on the car will cause the mercury to climb the walls of the relevant switch and so close the circuit to give warning to reduce speed.

A similar device l built for my "wagon" has proved to me just how easy it is to be lulled into a false sense of security and exceed the limitations of the vehicle.


Fig. 2

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## that Special problem

continued from page 545
and advertisements suggest that this is wise. Bearing in mind the volatile state of the electronic component market (and the inevitable lapse between going to press and publication) it would be hardly surprising if we slipped up on occasion. With regard to the more exclusive type of components, we always verify that they are listed either in the manufacturer's or a distributor's catalogue. It remains for the retailer to make his own arrangements to procure such items.
A further suggestion that we should restrict our contributor-designers to "popular" components is quite untenable: it is defeatist and unrealistic. Any magazine that has pretensions to be known as serious and forward-looking must at times venture beyond the wares displayed by the average retailer. And indeed how do components become "popular"? Often because of the attention originally drawn to them through some interesting and appealing project published by a magazine. Public demand then encourages traders to locate a source of supply and to incorporate such items in their catalogues and lists.

Component availability is a big and important subject. It involves a number of different trading organisations operating at different levels. The market they endeavour to serve is composed of a vast number of individuals, whose requirements are bound to be diverse in the extreme. Some contentious issues emerge when this matter is discussed openly, but we must all try to appreciate the difficulties and problems of others involved in this business.

Finally, although it is but little consolation, it is perhaps worthwhile remembering that industrial users have similar difficulties and often have to endure long delays for some components.
F.E.B.

A representative selection of letters from the large correspon.dence received on this topic appears in Readout this month. page 581

## 

We very much regret that Back Numbers of Practical electronics can no longer be supplied. Consequently, it is now more important than ever to place an advanced order with your newsagent to make sure of getting your copy.
Alternatively by taking out a subscription order your own personal copy will be sent direct by post to your permanent address anywhere in the world.; Further details are given on the "Contents" page.


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This instrument will detect and graph earth tremors produced by natural or man-made causes.
Consisting of a seismometer, pen amplifier and pen recorder, this inexpensive device is well within the capabilities of any enthusiastic amateur to build and should prove particularly useful as a group science project for schools

## PRACTICAL

ELECTRONICS


A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought.
This is YOUR page and any idea published will be awarded payment according to its merit.

## TEMPERATURECONTROLLER

THE temperature controller circuit shown (see Fig. 1) was initially designed for keeping cactus seeds at a constant 40 degrees Fahrenheit in a greenhouse. However, when the circuit was constructed it was found that by adjusting the potentiometer VR2 it would keep a room or greenhouse at a constant temperature from 30 degrees to 90 degrees Fahrenheit, $\pm 3$ degrees.


Fig. I. Circuit diagram of the temperature controller. The thermistors X1 and X2 are types VA1070 as used in television heater chains and are rated at 0.3A

The combined resistance of R1 and VR1 gives fine control over the forward bias of TR1. It was found in practice that TR1 continued to function perfectly with no thermal runaway component in the circuit; even in the hottest condition.

The relay was the only one available at the time, its working voltage was 20 V , although it pulled in at 7.5 V and the coil resistance was 1.5 kilohm. A relay of a lower working voltage would obviously improve the performance of the circuit.

When setting up the controller, the end of VR2 connected to the collector of TR1 should be disconnected and a voltmeter connected between the collector and the positive line. VR1 should be adjusted to give a meter reading of 7.5 V . When VR1 has been adjusted to give the correct setting the meter should be removed and VR2 reconnected -the controller is now ready for use.

It should not be difficult to replace the relay with a thyristor. This should be done by omitting RLA1 and D1, replacing this with a 1.5 kilohm resistor. The collector of TR2 could then operate the gate of a thyristor to switch on the heater. This has not been tried in practice.
A. D. Huff,

Dagenham,
Essex.

## POWER SUPPLY CUT-OUT

「N transistor power supplies there often arises a need for a simple overload cut-out circuit. The one shown in the circuit diagram Fig. 2 uses a minimum number of components, is quite fast-acting and is extremely versatile in application.
The resistor R 1 sets the limiting current, which can be anything from a few milliamps to the maximum current rating of the relay contacts. When the limiting current is reached, the transistor is turned on by the voltage across R 1 and the relay contacts open, thus isolating the power supply from its load and at the same time providing a self-latching effect which drives the transistor even harder into conduction. The circuit can be reset by a push-button switch across the relay contacts. An overload indicator lamp LP1 may be used if a relay with changeover contacts is incorporated in the device.

No component values are given for R1, TR1 and the relay as these will depend on the individual power supply. The transistor should be silicon; adequately rated to withstand the current requirements of the relay and the voltage of the power supply. For supplies using a negative earth a $p n p$ transistor is used, and D1 is reversed.

The relay should have a switching voltage about 5 volts less than the supply voltage.
The value of R1 can be found approximately from the following formula

$$
\mathbf{R 1}=\frac{0.6}{\text { Limiting current }} \text { Ohms }
$$

The value is, however, best found by experiment after using the above formula to find the approximate value.
G. H. Birkinshaw,

Brimington,
Chesterfield.


Fig. 2. Circuit diagram of the power supply cut-out

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Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned.

## EXPERIMENTER'S BOARD

The Chequerboard system by Circuit Integration Ltd comprises a series of patchboards and component carriers designed to facilitate the design, construction and proving of logic systems.
The patchboards, which can accommodate from 12 up to 4814 pin carriers according to size, consist of a series of sockets on $\frac{1}{4}$ in centres into which the carriers are plugged. The arrangement of these sockets is such that each pin of the carrier has two further sockets connected to it, so that interconnections can be made with the patchcords provided.
plain carrier enables units to be constructed from discrete components. A very good point is that provision is made on each carrier for decoupling the supply lines.

The system also includes a number of pre-wired modules which include a switch unit, a pulse generator, an indicator unit with three filament lamps, and several TTL gates and flip-flops. It would perhaps be less confusing if the indicator unit followed the normal positive logic convention, i.e. lamp ON for a positive, rather than a zero, input level.

The complete system has been well designed and certainly makes the wiring-up of a prototype logic circuit a very simple and rapid operation. The totally enclosed and very rugged construction will commend it to industrial and educational applications, although the price will probably put it out of the reach of all but the most dedicated amateur.

Full details and price list for the Chequerboard system is available from Circuit Integration Lid., Canal Street, Runcorn, Cheshire.

## SUPER I.C.

After two highly successful years, the. Sinclair IC-10 integrated circuit

With the addition of very few external resistors and capacitors the Super IC-12 makes a complete high fidelity audio amplifier suitable for use with pick-up or f.m. tuner. For more elaborate systems it can be used in conjunction with modules from the Project 60 Range.

Each Super IC-12 is supplied with a comprehensive Manual giving full circuit and wiring diagrams for numerous applications including high fidelity amplifiers. car radios, and oscillators.

The price of the Super IC-12 is $£ 2.98$ including a printed circuit board for mounting.

## STYLUS BALANCE

Calibrated in $\frac{1}{4}$ gram divisions, the Bib Model 32 stylus balance is claimed to be the first balance designed for determining the pressures of modern cartridges.
It has a non-magnetic base mounted on foam plastic and the crossbar of the beam has recesses which are mounted on a pair of pivot points. The end of the pivot boom has a red rubber insert on to which the pick-up stylus has to be lowered.

The price of the Bib model 32 stylus balance is $£ 1.80$ and as the balance will only be used occasionally, it is packed in a robust plastics case.


Power is conveyed to each carrier through two special pins which engage in further rows of bussed sockets, placed so that the carriers cannot be inserted the wrong way round. The power busses are wired to sockets on the side of the patchboard so that connection may be made to a suitable power supply.

Although the emphasis is clearly on digital circuits, one of the patchboards has an extra row of bussed sockets which allows for the provision of the $+, 0,-$ supplies required by many linear circuits.

A *wide variety of 14 and 16-pin carriers is available. The 14 and 16 -pin DIL, 8 and 10 -pin TO5, and flat-pack i.c.'s can all be accommodated, either by soldering to a blank carrier or, more usefully, by plugging into carriers which have the appropriate sockets on them. A


1C-12 marketed by Sinclair :Radionics


Bib Model 32 stylus balance from Multicore Solders
amplifier has had to give way to a successor.
Designated the Super IC-12 it has all the virtues of the original IC-10 but with the following improvements: greater output; lower quiescent consumption; works on any voltage from 6 to 28 V , without adjustment; full output into 3, 4, 5 or 8 ohms and a specially designed built-in heatsink.

The output claimed is 6 W r.m.s. continuous ( 12 watts peak). Frequency response is quoted as 5 Hz to $100 \mathrm{kHz} \pm 1 \mathrm{~dB}$. The total harmonic distortion is claimed to be less than 1 per cent (typically $0 \cdot 1$ per cent) at all output powers and all frequencies in the audio band. The quiescent current stated is 8 mA at 28 V . Input impedance is 250 kilohms nominal and the load impedance is 3 to 150 hms .

## LITERATURE

A comprehensive publication on Stroboscopes is available from Dawe Instruments. Written by W. V. Richings, the Company's Technical Director, it describes the basic principles of stroboscopy and the various stroboscopic devices, with particular emphasis on the modern electronic ștroboscope using a high-power discharge lamp.

The design criteria and technical features of stroboscopes are discussed and details are given of accessories and typical applications.

It provides a guide to the subject for technicians, production personnel and students. It is available free on request from: Dawe Instruments Ltd., Concord Road, Western Avenue London, W3 OSD.


Many home constructors must have required a fully variable, d.c. low voltage supply when experimenting with electronic circuits. One solution is, of course, to construct a fully variable, mains powered, stabilised power supply. However, there are the odd number of occasions when a mains supply is not available and a battery is the only source of power. With this in mind, the author has designed a self-powered, electronic, voltage stabilising control unit, the design and construction details of which are given in this article.

The stabiliser can accept input voltages of up to 85 volts d.c. (depending on the output voltage setting) and provide a fully variable stabilised output voltage from 0 to 30 volts d.c. at up to 250 mA
load current. To obtain a stabilised output voltage the input voltage must exceed the required level by 2 volts. The stabiliser draws virtually no current from the external supply and is completely selfcontained. Automatic current limiting is built into the unit and can be set to any one of five values, namely, $10,25,50,100$ or 250 mA .

By connecting a resistor across the output terminals, a constant current is available between the external voltage supply and the input terminals of the stabiliser. The value of the constant current is determined by the value of the load resistor and the output voltage setting of the unit. It can be set to any value from $250 \mu \mathrm{~A}$ to 250 mA .


Fig. I. Block diagram showing fundamental stabiliser

## PRINCIPLE OF OPERATION

The complete circuit diagram of the unit is shown in Fig. 2. In order to understand the operation of the circuit it is shown in simplified, block diagratmatic form in Fig. 1. The output voltage of the unit is determined by the values of $V_{1}, R_{\mathrm{k}}$, and $R_{\mathrm{v}}$, and is equal to, $V_{\mathrm{o}}=V_{\mathrm{r}} . R_{\mathrm{V}} / R_{\mathrm{k}}$, since both inputs to the voltage differential amplifier will be zero volts when the output is stabilised and thus, $I_{v}=V_{0} / R_{4}$ $=V_{\mathrm{r}} / R_{\mathrm{h}}$. The output voltage, $V_{0}$, is directly proportional to $R_{\text {, }}$ if both $V_{r}$ and $R_{\mathrm{k}}$ are kept constant. Keeping $V_{r}$ and $R_{k}$ constant has the advantage of maintaining the sensing current, $I_{\sqrt{ }}$, constant for all values of the output voltage.


## SPECIFICATION . . .

Stabilised Voltage Output: 0 to 30 volts in switched steps of 2.5 volts, plus 0 to 2.5 volts, fully variable. Input voltage must be between 2 and 55 V in excess of output level required.
Constant Current Capability: $250 \mu \mathrm{~A}$ to 250 mA , continuously variable.
Current Limiting: Five ranges: $10,25,50,100$ and 250 mA .

Accuracy of Output Setting: I per cent on all switched voltage ranges.

## Stabilised Voltage Stability:

(a) $\pm 0.1$ per cent for variations in the input voltage of from 2 to 35 volts in excess of the output voltage.
(b) $\pm 0.1$ per cent for variations in the external load from 0 to 250 mA .
(c) $\pm$ I per cent for normal room temperature variations
(d) $\pm 0.1$ per cent for variations of $\pm 10$ per cent of the unit's internal battery supply voltages.
Output Impedance: Less than 0.03 ohms over audio frequency range (No current limiting).
Ripple Voltage: Less than 2 mV peak-to-peak on full load (when used with a full-wave rectified unregulated supply of 35 volts output).

Providing the output current is below its limiting value, the output voltage is stabilised and the current flowing into the base of TR3 is controlled by the voltage differential amplifier. (Both the voltage and current differential amplifiers are in fact voltage differential amplifiers; the words "voltage" and "current" are used to describe the function that the corresponding differential amplifiers control, i.e. the "voltage" differential amplifier controls the output voltage and the "current" differential amplifier controls the limiting current).

When the output current nears its limiting value, the voltage developed across $R_{t}$. nears the reference voltage level fed to the current differential amplifier. which is equal to $V_{r} \cdot R 2 /(R 1+R 2)$. The result of this is that current is drawn away from the base of TR3 by the current differential amplifier thus reducing the output voltage in order to maintain a consiant current through $R_{c}$.

When completely current limiting the current differential amplifier has full control over the current flowing into the base of TR3, the voltage across $R_{\text {c }}$. is stabilised at a value equal to $V_{1} . \mathrm{R} 2 /(\mathrm{R} \mid+\mathrm{R} 2)$ and the current flowing through $R_{c}$ is maintained at a constant value equal to $V_{r}, \mathrm{R} 2 /(\mathrm{R} 1+\mathrm{R} 2) R_{\text {V }}$. Because the currents $I_{1,1}$ and $I_{\text {, are flowing through }}$ $R_{\text {, }}$. in addition to the load current, $I_{1}$, the output current in the limiting condition is given by the expression:

$$
I_{1}=V_{\mathrm{r}} . \mathrm{R} 2 / R_{\mathrm{t}} . .(\mathrm{R} 1+\mathrm{R} 2)-I_{1,1}-I_{\mathrm{t}}
$$

Now, the current flowing through $R_{r}$, is also equal to the emitter current of TR1, $I_{\text {"1 }}$, which is equal to $\left(\beta_{1}+1\right) . I_{1,1}$, where $\beta_{1}$ is the current gain of TRI. Thus $I_{1,1}=I_{, 11} /\left(\beta_{1}+1\right)$ and $I_{1}$ becomes:

$$
I_{1}=\beta_{1} \cdot V_{r} \cdot \mathrm{R} 2 / R_{c}(\mathrm{R} I+\mathrm{R} 2)\left(\beta_{1}+1\right)-I_{1}
$$

Thus, providing $\beta_{1}$ is very much greater than one, $I_{1}$ is approximately equal to:

$$
V_{\mathrm{r}} . \mathrm{R} 2 / R_{\mathrm{c}} .(\mathrm{R} 1+\mathrm{R} 2)-I
$$

Current $I_{v}$ has a maximum value of $V_{\mathrm{r}} / R_{\mathrm{k}}$ when the output voltage is stabilised (designed to be $250 \mu \mathrm{~A}$ ) and will have a value of approximately $V_{1} / R_{v}$ when the output voltage is reduced to a value $V_{1}$ due to current limiting.

## GENERAL CIRCUIT DESCRIPTION

Returning now to Fig. 2 it will be seen that transistors TR2 and TR3 form the voltage differential amplifier in a conventional long-tailed pair circuit. Transistors TR5 and TR6 form the current differential amplifier, again in a conventional long-tailed pair circuit. Matched pairs of 2N3707’s are used for the two differential amplifiers but virtually any pairs of $n p n$ silicon transistors with a current gain of greater than 100 at a collector current of $100 \mu \mathrm{~A}$ will do.

The common collector load of transistors TR3 and TR6 is made to appear very high by employing a constant current source provided by TR4. The emitter resistor of TR4 and the Zener diode, D1, determine the value of this constant current which is designed to be $100 \mu \mathrm{~A}$.

When the output voltage is stabilised (i.e, no current limiting) transistor TR6 is cut-off since its base voltage is below the value of the base voltage of TR5 with the result that all of the $100 \mu \mathrm{~A}$ constant
current, except that which flows into the base of TR7, flows through the collector of TR3. Any change in output voltage is inversely reflected at the base of TR2. A drop in the output voltage raises the voltage at the base of TR2 above zero and hence raises its emitter voltage. As the base of TR3 is connected to the zero volt line, by way of R20, a rise in its emitter voltage starts to cut the transistor off directing more of its collector current into the base of TR7. This in turn increases the current into the bases of.TR8 and the series regulator, TR9, reducing TR9's effective resistance and therefore the voltage dropped across it. The output voltage is thus restored to its original stable value.

When the current through the emitter resistor of TR9 reaches the limiting value the voltage at the base of TR6 equals or nears that at the base of TR5 with the result that TR6 starts to conduct, drawing some of the current away from the base of TR7. When this starts to happen the output voltage drops. TR3 starts to cut-off and the $100 \mu \mathrm{~A}$ constant current supplied by TR4 comes under the control of TR6. Any variation in the current flowing through the emitter resistor now controls the amount of current flowing into the base of TR7 and hence the current through TR9 emitter resistor. The output voltage is therefore automatically adjusted to maintain a constant current through TR9 emitter resistor.

## REFERENCE VOLTAGE

The reference voltage circuit consists of DI, TRI, R12, R13, and D2. The transistor, TR1, is fed with a stable voltage at its base by DI. The emitter resistor ( R 13 ) determines the values of the emitter and collector currents of TRI. Zener diode D2 is in the collector line of TRI and, since all the currents flowing away from the junction of TR1 and D2 are constant, the current through D2 will also be constant. Thus a stable reference voltage is maintained. Transistor TR1 is a silicon pnp transistor operating at a collector current of nearly 5 mA . The 2N4289 type specified is entirely suitable in this position, but any pnp silicon type with a current gain of the order of 100 at 5 mA collector current will be equally suitable.

The silicon diodes, D3 and D4, protect the base of TR2 from any large voltage swings. D2 also prevents the base of TR2 from going more than 0.6 volts positive when current limiting takes place. Any low current type of silicon diode is suitable in these positions.

## REMAINING COMPONENT FUNCTIONS

The $0.47 \mu \mathrm{~F}$ capacitor, C 2 , in parallel with the resistors around S2 and VR2, effectively increases the loop gain of the stabiliser to alternating voltages at the output, with the result that the ripple content is reduced as is the output impedance of the unit.

The $0.001 \mu \mathrm{~F}$ capacitor, C 3 , and the 22 kilohm resistor. R19, connected in series between the collector and base of TR3 reduce the loop gain of the amplifier at high frequencies and prevent instability that would otherwise occur.

The reference voltage for the current differential amplifier is derived from D2 by a potentiometer network consisting of R21, R22 and VR3. The current limiting ranges are set to their correct values by adjustment of VR3.

The resistor $R_{\mathrm{k}}$ referred to in Fig. 1 becomes R14 and VR1 in Fig. 2. The output voltage ranges are set accurately by adjustment of VR1.
The total consumption of the 9 V positive rail is approximately 11 mA . Two 4.5 V pocket lamp batteries (No. 1289) were chosen for this supply rail because of their large capacity and convenient size.
The total consumption of the 9 V negative rail is less than 1 mA . A 9V (PP3) battery was chosen for this supply rail because of its small size and adequate capacity.
A 4-pole, 3-way rotary switch (S1) switches the unit on and off. One pole (SId) is used to control the external supply and the remaining three poles

## COMPONENTS . . .

## Resistors

| RI-R | $110 k \Omega$-see text (11 off) |  |  |
| :---: | :---: | :---: | :---: |
| R12 | $1.1 \mathrm{k} \Omega$ | R23 | $2 \mathrm{k} \Omega$ |
| R13 | $510 \Omega$ | R24 | $16 \mathrm{k} \Omega$ |
| R14 | 16 kSz | R25 | $47 \mathrm{k} \Omega$ |
| R15 | $680 \Omega$ | R26 | $100 \mathrm{k} \Omega$ |
| R16 | 20k $\Omega$ | R27 | $100 \mathrm{k} \Omega$ |
| R17 | $43 \mathrm{k} \Omega$ | R28 | $4 \Omega$ (see text) |
| R18 | $24 \mathrm{k} \Omega$ | R29 | $10 \Omega$ |
| R19 | $22 \mathrm{k} \Omega$ | R30 | $20 \Omega$ |
| R20 | $680 \Omega$ | R31 | $40 \Omega$ (see text) |
| R21 | $16 \mathrm{k} \Omega$ | R32 | $100 \Omega$ |
| R22 | $5.1 \mathrm{k} \Omega$ | Rs se | e text |
| All $\pm 5 \%$ - ${ }_{6} \mathrm{~W}$, carbon high stabs. |  |  |  |

## Capacitors

$\mathrm{C!} 10 \mu \mathrm{~F}$ elect. 12 V
C2 $0.47 \mu \mathrm{~F}$ polyester
C3 $0.001 \mu \mathrm{~F}$ polyester
C4 $10 \mu \mathrm{~F}$ elect. 12 V
C5 $50 \mu \mathrm{~F}$ elect. 100 V
C6 $0.1 \mu \mathrm{~F}$ polyester 250 V
C7 $10 \mu \mathrm{Felect}$. 12 V

## Potentiometers

VRI $5 \mathrm{k} \Omega$ skeleton preset
VR2 $10 \mathrm{k} \Omega$ wire wound
VR3 $5 \mathrm{k} \Omega$ skeleton preset
Semiconductors
TRI 2N4289
TR2, 3 2N3707 (matched pair)
TR4 2N4058
TR5, 6 2N3707 (matched pair)
TR7 2N3707
TR8 BCI68
TR9 B504I
D! $3 \mathrm{~V}, 400 \mathrm{~mW}$ Zener diode
D2 $4.7 \mathrm{~V}, 400 \mathrm{~mW}$ Zener diode
D3, 4 any good low current silicon diode (2 off)

## Miscellaneous

MI 25 mA moving coil meter ( $2 \frac{1}{2}$ in diameter case)
SI 4 pole 3 way rotary switch
S2 Single pole 12 way rotary switch
S3 4 pole 5 way rotary switch (only 3 poles used)
SKI, SK4 Red screw terminals (2 off)
SK2, SK3 Black screw terminals ( 2 off)
BYI, BY2 No. $12894 \frac{1}{2} \mathrm{~V}$ battery (2 off)
BY2 PP3, 9 V battery
Veroboard $5 \frac{\operatorname{tin}}{} \times 2$ in $\times 0.15$ in matrix
Die-cast box $6 \frac{3}{4}$ in $\times 4 \frac{3}{2}$ in $\times 2 \frac{1}{8}$ in
4B.A. fixings, connecting wire, knobs for VR2, SI to S3, material for battery fixing.
connect the unit to the two internal batteries. In the first position of the switch all supplies are disconnected. In the second position the unit is switched on and in the third position the external supply is connected in addition to the internal batteries. This arrangement ensures that the unit is operating before the external supply is connected.

## RANGE SWITCHING

The sensing current, $I_{v}$ (see Fig. 1) is designed to be $250, \mathrm{~A}$, which gives a value for $R_{v}$ of 4 kilohms per volt. The maximum output from the unit is 30 volts giving a total resistance for $R_{v}$ of 120 kil o.'ms. A 10 kilohm wire-wound control (VR2) provides 0 to 2.5 volts fully variable and a twelve position rotary switch (S2), switching in successively
eleven 10 kilohm fixed resistors ( $\mathbf{R}$ | to $\mathbf{R} 11$ ), provides 0 to 27.5 volts in 2.5 volt steps. The rotary switch used for this operation is a single-pole 12-way wafer type. Eleven 10 kilohm fixed resistors are required for the construction of the switched ranges: 1 per cent, high-stability types could be used, but they are invariably large and expensive. The author found i: more convenient, and cheaper, to use $\frac{1}{3}$ watt, 5 per cent, high-stabs. of 11 kilohms in value and shunt them, as required, with 5 per cent resistors of values 100 kilohms, 200 kilohms, 390 kilohms and 820 kilohms. This method enables them to be matched to the value of the wire-wound control which is unlikely to be within 1 per cent of its nominal value.

## Next month: construction details




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IN Part I we considered the theoretical aspects of XEE's operational characteristics. This month, in the second and final part, we lobk at the system logic, construct the animal and perform the various tests.

## RANDOM CONTROL SYSTEM

Since our noise source produces spikes having a high occurrence rate, it is necessary to perform some processing before they can be usefully employed. The concept behind the scheme used here is quite a simple one (see, Fig. 6) which shows the basic principle). Noise pulses are fed continuously to the NAND gate, and the clock periodically "lets a few through" to operate the counter.

In the durations between clock pulses, the counter remains in the state previously set by the pulses occurring during the last clock period. It is at such times the counter can be "read" and so provide the required control information for the rest of the animal. It will be seen now that the occurrence rate of the noise pulses in no way affects the rate at which the random control data appears.

The principle is thus one of "throw the dice-look at the result-use the information", and so on. Indeed, a set-up of this type might usefully be employed in a dice throwing machine or, perhaps, a reaction-analyser.

## NOISE PROCESSING

The general idea is shown in Fig. 6, but in practice the system is a little different. In Fig. 5 we see that because JK flip-flops are used we are able to take advantage of the JK function for gating purposes. Hence the effect of the noise pulses at the clock input to the starboard control flip-flop will only be valid at times when the clock applies a logical $I$ at both the $J$ and $K$ inputs.

A clearer picture of what happens can be seen from the waveforms given in Fig. 7. From these waveforms it can be seen that during every clock pulse, or "window", the counter represented by the starboard and port flip-flops can be cycled many times before it finally comes to a stop. However, the important point to note is that this operation occurs quite randomly and hence the counter, following a clock period, can be set to any one of its severall possible states.


PART 2

> An animal approximation utilising integrated circuits to process optical and tactile sensing together with a random control to give reasonably lifelike responses


Fig. 7. Waveforms associated with random control system

The random control system has control, either directly or indirectly, over all gating functions within XEE. Since this information is derived from the $Q$ and $\bar{Q}$ terminals of the flip-flops, the control is complimentary.

## REVERSE GATING

The direction of rotation of the motors is dependent on the outputs at the starboard and port reverse gates. For forward rotation, the inputs to the particular gate must all be at logical 1 (output goes to 0 ). If both gates are in this condition, the animal will drive forward; if only one gate is producing a 0 . the animal will turn either right or left.

If any input to the gates is taken to logical 0 , the output will change to a 1 , with the result that the corresponding motor will reverse. Both gates in this condition result in reverse drive of the animal.

## INHIBIT REVERSE GATING

Inhibit reverse gating is provided by gates G2 and G3 (IC4); it is part of the stop function and serves to inhibit any reverse command given by the control flip-flops at such times. It is also included to ensure that correct homing on to light sources is provided when the optical sensors are stimulated (this only applies under conditions which will be mentioned later).

## STOP GATING

The stop function, overriding all other functions, is available through the agency of gate G3 (ICI). If both its inputs are at logical 1 then the output is at 0 , with the result that all motor supplies are disconnected and the animal stops. If, on the other hand, either or both inputs are at 0 , then the output will be 1 , permitting resumption of motor operation. The stop gate also controls the inputs to the inhibit gates to ensure that under stop conditions all muscle control relays are de-energised and consequently current in this area is reduced to a minimum.


Fig. 8. Waveforms associated with load sensing and reversing bistable

## AVOIDANCE FUNCTION

Application of any load exceeding a given period of time or amplitude will result in the avoidance routine being elicited. This amounts to the animal reversing for a short while then reverting to whatever the random system has currently set. The routine might thus be: reverse and turn left, or right, or stop. Whenever the Schmitt trigger is fired, a negative-going pulse sets the reversing bistable which simultaneously applies a 0 to one input of the reverse gates. Both gates thus return a 1 to the muscle control circuits, and XEE moves backwards.

The duration of the reversing mode is determined by the time interval from when it began, to when the clock pulse arrives to reset the reversing bistable. This will always be random, and can never exceed a completc clock period; take a look at Fig 8, which indicates the type of relationships that can occur between clock and load sensing pulses.

For operation of the reversing bistable it is convenient to think of it being first reset by the clock. Since, at the clock pulse, a 0 is fed to G 2 (IC5) its output will go to logical 1. As a result, Gl (IC5) will have one input at the same level, and because its other input is connected to a capacitor ( Cl 1 ), this too is effectively at 1 ; the output from this gate is thus 0 . Due to the cross-coupling between the gates, G2 (IC5) will have one input held at 0 by the output from Gl (IC5). The bistable will remain in this state unless a pulse arrives from the Schmitt; if this occurs, a 0 will be effectively applied to one input of Gl (IC5) whose output will go to 1. As a consequence, both the inputs of G2 (IC5) will then be at logical 1 and its output will go to 0 . Again. due to the cross-coupling, this state will exist until the clock pulse arrives to reset the bistable.

## OPTICAL SENSE AND RANDOM FUNCTION

For simplicity, the optical sensors do not boast any lens system, abthough, of course, there is absolutely no reason why this kind of sophistication should not be included if the constructor wishes.

The way in which the optical sense operates will depend on whether direct sensing is used, or whether the constructor chooses to permit some degree of random control. Since operation of this section of XEE is more easily understood by reference to the direct sensing arrangement, we will consider this first.

## DIRECT SENSING

For this form of sensing, the connections to the input gating marked with an asterisk in Fig. 5, are disconnected. The gates, however, must be left connected in all other aspects otherwise the logic will be affected.

Under dark conditions, when neither of the sensors are illuminated, the animal will be under the control of the random system; it will be either turning in one direction or another, or driving forward. If, say, the starboard sensor is illuminated, a logical 0 will appear at the output of the starboard input gateing which will be applied to the inputs of the starboard reverse gate and the inhibit port reverse gate. The former gate will thus show an output of 1 and so cause motor MOI to go into reverse drive. At the same time, the inhibit port reverse gate's output will be 1 , and, provided the remaining inputs to the port reverse gate are also at 1 , the additional input will ensure that a 0 , and hence forward drive, is established for MO2.

This complementary control over both channels is necessary because, ostensibly, there could be a counter command from the random control system to the opposite channel, when, in fact, no optical input was present on that side.

A similar regime will be operative if only the other sensor is stimulated. If both channels are active, though, the effects of the inhibit gates will be nullified, since the reverse gates will each have a 0 on at least one of their inputs, causing them to both return outputs of 1 resulting in XEE driving backwards.

Simultaneous illumination of both sensors is fairly rare, but can be an obvious embarrassment because when it happens XEE will continue its backing routine (up walls, if need be!) until the source of light is removed. A way of overcoming this difficulty is discussed later.

An important aspect of the optical sense is that a form of homing function is permitted. Take the case where light has fallen on the starboard sensor; this will cause a turn to the right. In doing so, the machine will move this sensor away from the source of illumination, but this will also result in the port sensor being brought in to line with the source. If this occurs, the animal will turn left, and so on. Under these conditions XEE thus performs a kind of "serpentine" movement until it is fairly close to the light, when it will suddenly veer off to the left or sight.

## RANDOM CONTROL

If the inputs to G2 (IC1) and G3 (IC7) are left connected to the starboard and port control flip-flops,

the optical sense displays quite different characteristics. One interesting point is that XEE no longer shows quite the same zest for mounting walls when in the reverse mode!

Optical sensing under random control permits XEE to extricate itself from powerful light sources. This is achieved by means of the control over the input gates. As a consequence, even though an input may be present at either of the sensors, unless the relevant gate is in receipt of a 1 from its associated flip-flop, the input will be ineffective.

XEE can therefore (apparently) make up its own mind about what it does, and does not, wish to look at! This does of course mean that in this mode of operation XEE is now free of the fetters that most moths seem unable to shake off.

## CONSTRUCTION

An illustration of the general wiring scheme for the main circuit board is given in Fig. 9. This shows the wiring and layout on Veroboard. Additional, simpler boards which are involved with the muscle control circuits are shown in Fig. 10.

The main board must be drilled in accordance with the chassis holes shown in Fig. 11. The two smaller boards must be cut to the sizes given in the



Fig. 9. Layout and wiring of the circuit board
components list. All components are mounted by means of their leads; breaks in the copper strips "must be made before the associated components are fitted. It is most important to make sure that the complete width of the relevant copper strip is cut through.

Do not fix the photo-sensors until last since they, almost always, are blessed with thin leads which with little doubt will break off if "man-handled" too frequently!

Mounting of the integrated circuits is best done first; the job is not a particularly difficult one, but it is essential that care be taken to ensure that every one of the 1.4 leads in each I.C. go through the holes in the circuit board. Attention is drawn to this fact because it is easy to think that all the
leads are poking through the board, when in reality they are not. The trick, if there is one, in fitting the I.C.'s is to ease the seven leads on one side through first, then, with the aid of the three fingers of one hand placed in-line against the edge of the remaining pins, gently locate and press them through the relevant holes. As each I.C. is fitted in this way it is advisable to solder it in place, lest it falls out as the next one is being attached.

Wiring of the circuit boards, and interconnections between them, should be done with thin plastic covered wire. Since the pitch of the holes in the boards is small, the copper lands are necessarily close together and without care it is extremely easy to make accidental bridge-overs. Such errors should be looked for prior to connecting any supplies.


Fig. II. Chassis details



Fig. 10. General layout of XEE and muscle control circuit wiring


In case of difficulty in obtaining the specified relays, any miniature d.p.d.t. relay of the correct voltage, and having a resistance of more than 500 ohm, may be used; although it may prove necessary to mount substitute types on top of the chassis near the batteries.

## LOAD SENSOR

The load sensor, as we previously discussed, employs a reed switch. This is of the 1 inch variety and must have approximately 16 turns of 26 s.w.g. enamel or cotton covered wire wrapped around its middle. The winding should take the form of two eight-turn layers.

During the initial testing the coil will be firmly cemented to the glass envelope of the reed switch. but at this stage the free wire ends to the coil need only be gently twisted together to keep everything from unravelling. An illustration of the complete sensor is shown in Fig. 10.

## OPTICAL SENSORS

The two l.d.r.'s (X1 and X2) should be mounted so that when the board is in position they face forward. If a "body" is to be made for XEE these sensors can be incorporated in the front. Once the l.d.r.'s have been connected-up, an Araldite "fillet" should be made between them and the circuit-board. To ensure a good bond, the board should be roughened with a piece of emery paper just at the places of contact with the epoxy resin.

## INITIAL TESTING

Just before testing, give the boards a visual inspection to make absolutely certain that no dry joints. bridge-overs, or wiring errors exist. The muscle control boards will not need to be checked out now, but will be tested in conjunction with the relays later. Connect points " $b$ " and " $c$ " on the main circuitboard to an 18.0 volt d.c. source; " $b$ " must go to positive, and "c" to the negative. While performing the tests make sure to keep the optical sensors clear from any direct sunlight which could cause ambiguous results.

## RANDOM CONTROL SYSTEM

Connect a voltmeter (set to a range which will measure 600 V ) between earth and pin 13 of the starboard control fip-flop (IC2), and check that approximately every second or so there is a brief flicker shown by the meter. This will be the indication if the clock is functioning correctly. Now disconnect the lead from pin 13 and connect it to pin 10 (IC2). The meter reading should be either a steady 6.0 V (logical 1). or very nearly zero (logical 0); this indication should be interrupted about every second by the effect of the clock and noise pulses. It is important that this last reading changes in a random fashion periodically. Next, disconnect the lead from pin 10 and reconnect it to pin 10 on the port flipflop (IC3). The reading should be similar to the last, to wit, it must change randomly from 1 to 0 every now and then. If no change is observed, then connect the meter between earth and pin 3 of the starboard flip-flop (IC2). In this position it should indicate a regular flickering due to the noise source. If this is not so, the Zener diode D1 should be disconnected and replaced by another one. Sometimes one comes across a particularly "un-noisy" diode, these "good ones" are however no good to us!

## STOP AND REVERSE GATING

Connect the meter between earth and the output of the stop gate. This should periodically change its logical state from 1 to 0 , or from 0 to 1 . Disconnect
the meter from the stop gate and reconnect it with the output of the inhibit starboard reverse gate. Now disconnect the leads going to pins 5 on the two control flip-flops, and apply the light from a torch to the port optical sensor.

The meter should indicate a logical 1 condition as long as the light remains on. Repeat this test with the meter connected to the output of the inhibit port reverse gate, but with the light applied to the starboard optical sensor. Again the meter should indicate the 1 state all the time that the light remains on. With the light still on, connect the meter to the output of the starboard reverse gate. The meter should again show the logical I state. Reconnect the meter with the output of the port reverse gate. This too should be at 1 when the light is transferred to the port sensor. Leave the meter connected for the next test.

## REVERSING FUNCTION

With no light applied to the optical sensors, trigger the reversing bistable by momentarily touching an earth connection on either one of the inputs of G1, IC5 (pins 12 and 13). The meter should immediately indicate a logical 1 condition (if it is not already). This reading should eventually change once the reversing bistable has been reset by the clock pulse and the system returns to the control of the random control section. Connect the meter to the output of the starboard reverse gate. Again trigger the reversing bistable and ensure that the results are the same. Disconnect the meter and the 18.0 volt supply. Reconnect the leads to pins 5 of the control flip-flops.

## CHASSIS CONSTRUCTION

The type of material from which the chassis is fabricated is not particularly critical; for the prototype, 16 gauge aluminium was used. The chassis takes the form of a single piece of material having two $\frac{3}{4}$ inch right-angle sections folded down the length of its sides. The rear side plates carrying the gear shafts and wheels are mounted on this chassis.

A "U" bracket, situated on the chassis midway between the two side plates, serves to support the inner ends of the gear shafts. A further " $U$ " bracket is utilised for the frontally located castor wheel. All metalwork should be constructed according to Fig. 11. Any bending which is required can be done with the aid of two small pieces of tough wooden plank and a vice.

## MOTORS, GEARS AND WHEELS

As Fig. 10 shows, both motors are situated side-by-side towards the rear of the chassis. Each motor has a friction-fit worm attached to its output shaft, and this is arranged to drive a further shaftyat right-angles to it by way of a wormwheel. The gear ratio thus provided is $40: 1$. This lay-shaft is also fitted with a pinion which, in conjunction with another pinion, on the final drive shaft, results in a further reduction of $3: 1$. In this way an overall ratio of $120: 1$ is achieved.

The rear wheels, one attached to each final-drive shaft, are $1 \frac{1}{2}$ inch diameter soft rubber aircraft type. These should have their centres drilled out a little to accommodate $3^{3}$, , inch shafting, and be secured in position with epoxy resin. In order to achiever improved traction, the rear wheels should have their

tyres "scalloped" around the circumference. This is best performed using a pair of small side-cutters, pinching-out the required amount of rubber and nipping it off. The resulting tread is fairly coarse, but serves its purpose, since, on nylon carpet, the wheels would undoubtedly slip.
The castor wheel is the same as the type used at the rear but, naturally, without the tread. The wheel is fitted to a shaft going through offset holes in the "U" bracket which swivels on a 4B.A. screw at the front of the animal.
Shaft retention is obtained either through the use of plastic tubing, or by means of washers soldered to the ends of the shafting. Sometimes the nylon gear-pinions have casting flashes still attached to them; this is particularly noticeable on teeth. The flashes must be removed with a sharp knife, or razor-blade, before the pinions can mesh properly.

## MOUNTINGS

Four 4B.A. screws secure the main board to the chassis. The screws are initially bolted to the chassis, their nuts serving as spacers to separate the board from the metal body. Once the screws and nuts have been fitted, a layer of insulation tape, approximately the same size as the main board, should be placed directly beneath where the board will be located. This will ensure that no exposed wiring gets shorted out on the chassis. At this stage the supply leads,
outputs $w, x$ and $y$ and the connections to RLA should be connected up and have sufficient length to reach the other areas of the animal to which they will be later connected. The main board can now be fitted.

Relays RLB and RLC are mounted one above the other, secured on a pair of long 6B.A. screws. The Veroboard associated with these relays is mounted on RLB. The board is fixed in place with an impact adhesive.

The remaining relay, RLD, is kept in position with this adhesive, as is its associated component board.

## SYSTEM INTERCONNECTIONS

The various interconnecting wires are shown in Fig. 10. Once the relays have been connected to their respective boards, the inter-circuit wiring can be completed. The latter, when finished, should be neatly laced up or cleated so that all leads are formed into one common cable-loom. The free ends of the leads going to the batteries should be passed through the grommet in the chassis and terminated in crocodile clips. Constructors may, at this point, choose to include a switch for isolation of the batteries from the rest of the circuit; for both simplicity and minimum cost, however, it was considered sufficient to employ crocodile clips for the purpose at the time.

Before connecting any batteries, do make absolutely certain that there are no wiring mistakes.


Fig. 12. Battery fixing and wiring details

## BATTERIES

The motors are powered by a pair of type HPII batteries. These should be strapped together according to Fig. $12 \cdot$ using masking tape or similar material. A link between the positive of one battery and the negative of the other should next be soldered into position. The free ends of the batteries will be available for connection to the motors.

The two PP6 batteries, that provide the circuit supply, should undergo a similar process (see Fig. 12). The batteries are mounted on top of the chassis directly above the motors.

## FINAL TESTING

Connect up the motor supplies first; note that motors do not run. If this is not so, check wiring from relays. Temporarily disconnect the motor supplies if all is satisfactory. Connect the supplies to the logic and muscle control electronics; note that a regular ticking sound is evident from the RLB, RLC and RLD. If this is the case, then reconnect the motor supplies.

With both supplies connected the rear wheels should be either rotating so as to give forward motion of the animal if placed on the floor, or rotating in opposite directions. On no account should the rotation be that which would result in reverse motion. If, occasionally, this reverse action does occur, then disconnect the outer end of RLA operating coil and remove one turn. Reconnect the coil and try again. Repeat the process until conditions are just stable, i.e. motors not going into the reverse state when unloaded.

Apply a load to the rear wheels by attempting to stop them with the fingers; this should be done for a period of at least I second, or at least until reverse motion of the wheels occur. If no reverse action is noted short out the contacts of RLA and ensure that reverse action occurs. If this is not so, then check the Schmitt trigger or reversing bistable. Assuming that XEE reverses when RLA contacts are shorted out, then add one turn to the coil and test the function again. Once this function is satisfactory, the coil can be cemented in place. Shine a torch on to the optical sensors, whilst XEE is moving forward, and establish that the motors respond by reversing for the opposite side to that stimulated; this need not happen straight away since the random
function is also involved. The animal can now be placed right way up and begin its life proper, journeying around the chair and table legs in your living room.

## BODY

The constructor may wish to construct a shell for XEE to give it a more animal like appearance. The shell shown on last month's cover and in the various photographs was constructed by inflating a balloon to the required size and covering it with tissue paper or newspaper torn in small pieces and pasted with polycell glue.

The shell is built up layer by layer until it is about ${ }^{1} 16$ inch thick; it is then left to dry fully before slowly deflating the balloon. The bottom edge of the shell is then trimmed and a thick cardboard base cut to shape so that it fits inside the shell about 1 inch away from the bottom edge. This base is then fixed to the shell using more paper and paste and finally a hole is cut from the centre to accommodate the batteries and XEE is secured to the base.

Decoration can be painted on the shell as desired and eyes made from the plastic lamp covers glued on over holes in the shell behind which can be mounted X1 and X2. The prototype XEE utilises a selection of resistors to form a spine and has an on-off switch as a tail!

## CONCLUSION

XEE is only a partially non-deterministic animal. That is to say it does not always respond to stimuli with any predictability. However, despite this, the beast does stand a greater chance of "survival" than, say, one which acts with 100 per cent certainty. Nevertheless, this in no way implies that it has any intelligence, although it does have the chance of reacting as if it did.

## ACKNOWLEDGEMENT

The author wishes to acknowledge assistance given by J. Salmon in connection with the mechanical aspects of XEE; also J. D. Pountney who devised and constructed XEE's "body".


THE painting of anything made of metal, particularly a ferrous metal, that is exposed to a marinc atmosphere, becomes a process requiring somewhat more than ordinary care.

Air charged with a saline content soon corrodes away metal work not properly protected. Objects immersed in the sea itself, are subject to a great rate of corresion, this being accelerated by electrolysis acting between dissimilar metals that are touching or in very close proximity to one another.

Even painting the hull of a boat, and we are here concerned only with metal hulls, does not in itself guarantee an indefinite degree of protection, since any imperfections in the paintwork will allow the salt water to come in contact with the metal hull and begin its corrosive work.
wats riddled with holes, the existence of these not being known prior to the tests.

## RESISTANCE METER

The detector is essentially a resistance meter that measures the resistance occurring via the pin hole, between the metal and a probe that is passed over the paint work.

This probe, connected to the input of the detector, consists of a length of broom handle to the end of which is clamped a piece of sponge material containing the search wire. In use, the sponge is soaked in water with the addition of a wetting agent. As the probe is passed over the paintwork, the water, its surface tension reduced by the wetting agent, flows into the pin holes and combines the circuit between the metal and the probe.

## A PAINT PIN-HOLE DETECTOR



Visual inspection of the paintwork becomes at boring and laborious process, and will not reveal pin prick holes and hair line cracks that will permit the ingress of the sea water.

## STANDARD COMPARISON

The device that forms the basis of the article is cheap, easy to build and use, and has proved to be of real usefulness. Accurate and reliable testing of such at device is difficult, since "standards" in the form of graduated pin holes in a given thickness of paint, on metal, are not available. One therefore has to resort to comparison with commercial pin hole detectors, and this was the standard used when evaluating the prototype.

The commercial detector utilised a radio frequency generator providing a low current output. at a voltage up to 25 kilovolts. The test piece was the metal partitioning in a block of offices, that visually appeared to be in very good condition. An area was marked off and tests commenced, the home made detector performing first.

The results were extremely pleasing as the commercial detector agreed with the home made detector every time.

It was a revelation to discover that the paintwork

## CIRCUIT ACTION

The circuit of the detector is shown in Fig. 1, and will be saen to be very simple. Transistor TR 1 comprises the input stage, TR2 is a simple CR audio frequency oscillator, and TR3 is the output stage feading a 250 ohm magnetic carpiece.

Under normal circumstances, the batse of TRI will be floating, and only a small leakage current will flow from collector to emitter.

Whon TRI is cut off, the leakage current is insufficient to bias TR2 on and it will not oscillate. If the resistance between the base of TR1, and the positive supply line (between which the probe is connected) is such as to allow TR1 to pass sufficient current to bias TR2 on, it will oscillate and the earphone in TR3 collector will emit a whistle.

If the probe is short circuited, TRI will be tottomed, allowing approximately 8 V to appear between TR1 emitter and the negative line.

For this particular application, all we want is a noise to signify pin holes, we don't want $\mathrm{Hi}-\mathrm{Fi}$, so TR3 is a very primitive output stage which does provide a very lively whistle when the occasion arises.

The current drawn from a 9 V battery is very modest. 2.5 mA at quiescent, rising to 3.7 mA maximum with the probe short circuited.

## CONSTRUCTION

The prototype was made up on a piece of plain $0 \cdot 1$ in pitch peg board, shown in Fig. 2. The compenents are inserted from one side, and the leads interconnected on the other side, sleeving being used where appropriate. Any plastic box with dimensions approximating 3 in $\times 3$ in $\times \frac{3}{4}$ in will provide a suitable housing.

Fer maximum sensitivity, the transistors should be as specified. Experiments have shown that TRI can be successfully replaced by a BC109C, and TR2/TR3 by either a BC109 or a BC108.

Depending on components used, R1 may require an alteration in value. With the prototype, a resistance value was found such that vigorous oscillation occurred with the probe short cireuited, with a still useful signal available when the probe was replaced


Fig. 2. Board assembly and wiring details
by a string of 10 megohm resistors in series providing a total value of 100 megohms. Construction ef the probe is simple as shown in Fig. 3.

When the prototype was completed and performing satisfactorily, the circuit board was isolated from the rest of the components and then encapsulated with a silicone rubber compound, so completely water-proofing it. This is very desirable, if not essential, in equipment that is used with water.

## USING THE DETECTOR

In use, the positive line is clipped to the painted metal, ensuring a good electrical connection. The length of line required will clearly depend upon individual circumstances.
The probe is then dipped into a solution of water and wetting agent (a proper wetting agent is definitaly helpful and should be used whenever possible. The use of a saline additive does improve conductivity; however, it may get into holes and cracks and start the very corrosion we are trying to prevent, so its use is not recommended) and passed over the paintwork in slow regular sweeps. As soon as the probe's sponge loses its water it must be recharged.

A steady whistle will indicate a single pin hole, whilst a bleep or series of bleeps will indicate a singlo pin hole, or a whole colony of pin holes. The suspect


## COMPONENTS . . .

| Resistors |  |  |  |
| :--- | :--- | :--- | :--- |
| RI | 270 k | R4 | 4.7 k |
| R2 | 3.9 k | R5 | 560 k |
| R3 4.7 k |  |  |  |
| all $10^{\circ}{ }^{\circ} \frac{1}{8} \mathrm{~W}$ |  |  |  |

Capacitors
CI-C4 0.01 polyester ( 4 off )
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## Miscellaneous

PLI, 2 Miniature plugs (2 off)
SKI, 2 Miniature sockets (2 off)
SI Single pole on/off slide
BYI 9V PF3
Plastics case 3 in $\times 2$ in $\times \frac{3}{4}$ in


Fig. 3. Probe assembly details

area is then circled with chalk-not grease pencils or "Chinagraphs". as these may show through when patinted over-for subsequent repainting.

## SPURIOUS SIGNALS

As TRI base is, in effect, floating, it is possible for an electronic charge to build up between the base probe and adjacent components. This charge will cause random bleeps when the probe is used in the described manner, causing the operator to erroneously suspect the presence of non existent flaws in the paintwork.

A high value resistor, 10 megohms or so, connected between base and the negative line will prevent the building up of such static. This addition may lower the sensitivity though it should not be so low as to affect the detector's usefulness.

The prototype was designed to be slipped into an overall top pocket, with wires terminated in wander plugs permanently attached to the probe's broom handle. The case can be attached to the handle if required, providing a single composite assembly.



It was nice to renew a visual acquaintance at least of Ithe magnificent Willis concert organ in the auditorium of Alexandra Palace, the venue of the fifty-fifth Physics Exhibition held during the period April 19-22.

The setting provided by this 19th-century masterpiece well suits the academic and scientific tenor represented by universities, research establishments and industry.

Representation by these various factions provided a full hall of exhibits, the mrost conspicuous of all being the large central stand organised by the ltalian Federation of Scientific and Technical Associations and other bodies. Besides providing a showcase of current developments in instruments and apparatus, there was also a large selection devoted to the history of physics in ltaly with career cameos of some of the more distinguished physicists.
Of the exhibits, probably one of the most exciting was the Mullard Research Laboratory's demonstration of microwave television broadcasting using an intermediary satellite transmitter and standard colour television set for reception.
The implications of this are tremendous, as a satellite aerial 20,000 miles above the earth can cover a whole sub-continent.
Complex and expensive ground networks are not required which means that a developing country can be equipped with a service in a fraction of the time necessary to install a conventional television system.

For technical reasons. satellite television systems in advanced countries would probably use the frequency band 11.7 GHz to 12.7 GHz with either digital or frequency modulation. Commercial success would depend greatly on the availability of cheap, reliable microwave components.

A solid state digital clock with an unusually small readout display proved an arresting item of the Marconi stand. This indicated hours and minutes using four characters each of which consisted of seven tiny electroluminescent bars providing a total character size of 0.185 in by 0.125 in . (!)

An accompanying exhibit demonstrated the display capabilities of organic nematic crystal materials under the action of an electric field. Requiring only low power, six etched alpha-numeric characters made up on plates of conducting glass containing a thin film of liquid crystal material, were illuminated by a simple keyboard switching.

The advantages of these display elements are low running costs and good resolution and contrast under conditions of high ambient lighting. Perhaps in these materials, we will see the demise of advertising's neon jungles.

While these display examples represent the research of companies there was a great deal to be seen which will almost certainly be the object of commercial development.

Measurement devices always figure strongly at this exhibition, and this year proved no exception. In this department the Ministry of Defence showed a corrosion monitor for measurement of corrosion rates. Corrosion is a subject of very great technical importance that is still not fully understood and is the subject of research of a number of laboratories.
Every refinement in instrument technique pushes back the frontier of scientific ignorance and it is at this exhibition that this is most manifest.

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Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

## Tiresome and uneconomic

Sir-Replying to your comments regarding specialised components has been very difficult. I have tried to offer some constructive suggestions. First, however, I am thoroughly convinced that the onus lies with the various authors writing in your magazine and that your editorial staff should ensure that the specialised components needed for a particular circuit are available to the retail customer.

With respect, I would quote your small independent competitor, Radio Constructor, who for years have made a point of advising us in advance that such and such a transistor would be specified. Also we are informed of the approximate publication date and then lay on stock to meet the demand. I have repeatedly suggested this approach to you and to your staff.

1 dispute your suggestion that specialised components are not available to the retail public. For years we have supplied specials. However, I would state that this has been done as a service to the customer and has usually resulted in a thoroughly uneconomic transaction. In the last few days I have discussed this matter with several other suppliers of repute and we are all of the opinion that specials are a tiresome and uneconomic matter but which we put up with as a service.

I could go on to stress many examples of costings done on the economics of obtaining special parts for customers, but I think I have said enough on the subject to convince you.

In attempting to make a constructive suggestion, I have thought along the lines of charging a $\ddagger 1$ service charge for specials, but the fact is that the enthusiast of today is primarily experimenting for his own enjoyment and has a limited budget, and so such suggestions may not be a proposition.

My company is now distributing the products of Siemens, Newmarket, Marston Excelsior, Schauer semiconductors and many others. and we carry very much larger and extensive stocks of components than we offer for retail.

Our mail order operation has now been carefully examined and we are offering a wide range of popular components at competitive prices and with excellent delivery. and I know beyond any shadow of doubt, merely from the tremendous response of the last month or so, that this is what the customer really wants. We will continue to try and assist customers with specials and very probably continue to make a loss on those specials. However, it is very few companies who will be prepared to do the same.
P. F. Clarke,

LST Electronic Components Ltd., Brentwood, Essex.

## Direct access

Sir-I was interested to read your editorial in the May edition of Practical Electronics where you discuss the problems encountered by the private constructor in obtaining direct access to components.

As you will no doubt realise from our advertisement in your magazine, we have started a mail order division which is specifically directed towards the requirements of the private constructor. Distronic Ltd. is an industrial distributor and a member of A.F.D.E.C. (Association of Franchised Distributors of Electronic Components). We are a very active industrial distributor and recognise the problems encountered by your readers. For this reason we are attempting to do something about it and are hopeful of being able to offer the same service to private individuals as we are currently offering to our industrial customers.

The products we are offering are all supplied by leading manufacturers under guarantee, and we are able to pass on to the private constructor some of the cost advantage we have gained by buying in bulk.

Looking forward to a long and fruitful association with your magazine.
M. H. Jacobs,

Distronic Ltd.,
Harlow, Essex.

## Workuble solution

Sir-I have read the letters on the Readout page and also your editorial comment. As an active transmitting amateur I appreciate the problems facing your readers in obtaining specialised components, but the solution to this problem unfortunately lies with the distribution industry rather than the manufacturer. You will, 1 am sure, appreciate that it is quite impossible for a manufacturer to deal with individuals and in any case it is now usual practice for a minimum order charge to be levied to cover the high cost of order processing.

It would seem to me that the letter in your Readout page over the signature of Mr. A. Sproxton sets out a workable solution to the problem and 1 am sure that the usual distribution outlets will be quite willing to deal with such an organisation.
R. R. Adams.

Sales Manager,
Resistor Division,
Painton \& Co. Lid.

## Nol all roses!

Sir-We would like to make the following points which we believe are not understood by people outside this trade.

Firstly, the amateur market in the United Kingdom has seen a drastic expansion for its size over the last two years and individuals who have no connections with this industry have proven to be ambitious and professional in their approach to construction.

The quest for improving quality seems never ending, and the number of specialised components required for projects are becoming numerous and more expensive.

The second and perhaps the most important point we would like to bring to your attention is the faot that the trade has no warning of the components which are to be used in the various projects each month and in the case of Practical Electronics, readers will have to wait two months before they can see an advertisement showing the special components they may require. Also none of the trade are consulted by the authors of the articles, as to the price and ease of availability of components used.

The sad fact of this is obviously to reduce the numbers of potential constructors and, therefore, make many of the articles which are produced purely an academic exercise.

Finally, referring both to the letter published in the magazine by the private individual referring to the inability to obtain $I$ off, and your own comment, we must point

out that the costs involved both on the part of the manufacturers, or distributors and a concern such as ourselves, are somewhat prohibitive so that an article which would perhaps cost $£ 2$ to purchase would end up being sold by the retail customer at anything up to $£ 4$, if all concerned are to break even let alone make a profit. We believe that this is the major deterent to solving this problem in the manner suggested by your reader. However, we must commend his innovation to having personal order forms printed for the real enthusiasts.

Although this is an extremely long letter, it will only scratch the surface of the problem, but we hope that you will publish some of the content, so that the amateur enthusiasts will perhaps realise that all is not roses in this particular garden.
J. Marshall,
A. Marshall \& Son (London) Ltd.,

London, N.W.2.

## Distribution set-up

Sir-I have read with interest your Editorial concerning specialised components as well as the letters from your readers and can well appreciate the problem of obtaining specialised components which are not normally available from the usual retail outlets for components.

We ourselves are Specialised Distributors of electronic components and from time to time, we do get private individuals approaching us to obtain components. Wherever we can, we do try and help them and since it is normally a cash with order transaction, there is no problem in supplying one or two offs to private people and I think you will find that this is not peculiar to ourselves and quite a number of Distributors are quite willing to operate this way.

The problem is, however, in the private individual locating a source of supply for the component that he wants, bearing in mind that at the present time there are well over 80 Distributors of electronic components in this country, many of them specialising in a narrow field of items. For instance, ourselves, we specialise in the supply of Electrolytic Capacitors and Wirewound Potentiometers as well as a number of other items and in the main, you will find that this is the
pattern of Distributors throughout the United Kingdom.

There are, of course, exceptions to these and there are quite a number of very large Distributors indeed who handle a very wide range of components but because of their size it might well prove difficult for the private individual to purchase from these larger Distributors. Hence it follows that a knowledge of the distribution set-up is necessary if the person is to locate a source of supply for the type of specialised component he is looking for.

Publications are available which list all the Accredited Electronic Component Distributors in the United Kingdom, showing their location, contacts and the range of components that they handle. Using such a publication the private individual should have no great difficulty in obtaining the particular component that he wants.
D. E. Clarke, Intercontinental Components, Maidenhead, Berks.

## Customer relations

Sir-As you perhaps know, in general retail mail order a distributor carries a wide range of components in order that a customer may kit up his project in an order to one address. Certain projects require specialist components obtainable perhaps from only one source which is often indicated in an article. Thus in some cases two or perhaps three suppliers must be contacted. I feel that this situation must always exist since even mail order retail concerns such as mine must keep large quantities of "stock items" to be economical and satisfy the market and to keep small stocks of a wide range of items of uncertain demand is expensive on space, finance and organisation and results in an uncompetitive approach. I may say the problem is much more acute in industry where for a simple project several suppliers may have to be approached for good delivery of all components.
As a company we are quite prepared to order non-stock items from our normal suppliers for retail customers. In certain cases we advise against this and offer substitutes. There is a great problem here in that having ordered a component and given a delivery date as advised by the supplier, it is not uncommon to have the delivery date extended by three or even six months. This causes bad customer relations and although it is possible for us to forward cover our own stock several months ahead for known long delivery items, the retail customer is usually not prepared to suffer this inconvenience.

Until recently it has been our policy to offer components at a catalogue price and offer a discount to all customers, industrial and retail alike, the discount rate depending on order value. We have recently become industrial distributors or stockists for three companies and in addition to the discount arrangement we will be able to offer components at a lower (but net) price for quantities of the same item. This facility will be offered to retail customers on request. Thus we hope to extend our service and provide many i.tems formerly unobtainable.
D. A. Longland,

Electrovalue,
Englefield Green,
Egham, Surrey.

## Your responsibility

Sir-Regarding your leader in the May issue, 1 think you also have some responsibility regarding the marketing of components.

I recently started building the equipment for the Integrated Circuit Tape Recorder in the January 1968 issue, having had the deck for some time. I now have all the parts except the oscillator pot core, type LA2103. I visited a number of shops in London, and no-one can tell me where this can be obtained. Being retired I do not now have access to the information that might help me to use other cores in my possession.

If we cannot get components to build the equipment described in your magazine, there is little reason to buy your magazine.
H. Boys,

Weedon, Northampton.

## Mysiery of the Ferrites

Sir-With reference to your Editorial in the May 1971 issue, I am at present constructing the "P.E. Gemini", amplifier and having great difficulty in obtaining the ferrite components for it (FX2239 pot cores and DT2178 formers).
During the past year I have built a colour T.V. set from articles in another journal, and had the same difficulty in obtaining the many ferrite components for this, but after much phoning and letter writing 1 managed to obtain the parts from a number of suppliers. I have referred to these suppliers for components for the "P.E. Gemini" but with no success. Eventually, I phoned Mullard Lid. direct for assistance and was given a list of local distributors, I then got in touch with some of these and was told they could not trade directly with the public, and I would have to place my order with one of their retailers. When I called at the

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| No. of Pole | 2 way | 3 way | 4 way | 5 way | 6 way | 8 way |  |  |  |
| 1 pole | 60p | 60p | 60D | ${ }^{60 p}$ | 60p | 60p | 80 p | 60p | 60p |
| $\because$ poles | 60p | 60p | 60 p | 80 p | 60p | 60p | 60p | 11 | 11 |
| 3 poles | 60p | 60 D | 60 p | 60 p | 21 | 21 | 21 | 21.40 | 81.40 |
| $\pm$ poles | 60 p | 80 p | 60 p | $\underline{1}$ | 11 | 21 | 11 | 21.80 | 21.80 |
| $\therefore$ poles | 60p | 601 | E1 | 21 | 21.40 | 21.40 | 21.40 | 22.20 | 24.20 |
| ${ }_{5} \mathrm{i}$ poles | 60 p | e11 | E1 | 81 | \&1.40 | 21.40 | \$1.40 | 22.60 | 28.60 |
| 7 poles | 60p | E1 | E1 | £1.40 | 21.80 | 21.80 | 21.80 |  |  |
| ${ }^{8}$ poles | 21 | 21 | ¢1 | £1-40 | \&1.80 | 11.80 | 81.80 |  |  |
| 9 poles | \$1 | 81 | $\underline{11}$ | \$1.40 | 22.20 | 22.20 | 22.20 |  | 1 |
| 10 poles | 81 | \%1 | 11.40 | \$1.80 | 22.20 | 22.20 | 28.20 |  | , |
| j1 poles | \&1 | 21.40 $\$ 1.40$ | 1.40 $\$ 1.40$ | 21.80 21.80 | 42.60 22. 60 | 22.60 ¢2.60 | 22.60 22. 60 |  |  |

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impedance $13: \Omega$
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## J. BULL (ELECTRICAL) LTD. <br> Dept P.E., 7 Park Street, Croydon, CRO IYD

 Callers to Electronics (Croydon) Ltd., 102/3 Tamworth Rd., Croydon
retailers 1 was told, "Sorry, we do not sell spares". So back to where 1 started. 1 would be very grateful to you if you could inform me where these items may be obtained.

Whilst writing, I would like to mention that it would have been of assistance if you had stated in the components list where one could obtain such special items as the pot cores, 6 -pole 5 -way switch, etc. instead of resistors, capacitors and transistors, as these form the basic component stock of most retailers anyway.

In conclusion, I agree with your view that it is almost impossible to obtain special components. I can say that I have been deterred from embarking on some of the projects in the magazine for this reason.
A. J. Sanders,

Woking. Surrey.
You should be able to obtain all the Mullard Ferrox components for "P.E. Gemini" from Home Radio Ltd, Mitcham. The switches referred to were quoted as Maka types and are also available from Home Radio.

## No-one wants my order!

Sir-Since "Radiospares", often mentioned as a source of components, supply only trade, those people who are not in the trade can have great difficulty in obtaining such components.

For example, one of the transformers used in the "Digi-Clock" is quoted as available from them, and so far 1 have found nowhere else which will sell this type.

Moreover, since many places will not take single orders, I cannot order through them from "Radiospares". and I have therefore found it impossible to obtain such a transformer.
Can you suggest any remedy to the situation?
This problem does not only apply to the clock, but also to many other articles.
M. Easterfield, London, N.W.11.

## Makers' numes

Sir-Interest in your magazine has led me to try my hand at building some of the projects therein. As 1 am in a strange town and have difficulty in finding the parts, would you please let me know where I may get the kits specified in your projects. I am particularly
interested in the "Digi-Clock" project at the moment and would like that kit first. From that, of course, will lead many other efforts, and I hope with success.

In your components list I see several addresses bracketed (). Does this mean that these components may be got only from these people? e.g. Nixie numerical indicators (Electroniques (S.T.C.) Ltd., Edinburgh Way, Harlow, Essex) or (Midget Mains 250 V by Radiospares). Of Radiospares I could find no address and this is what led to the questions.

Thank you for a very interesting magazine, and may all your projects be successful.

> F. Lewis,

## Chaddesden, Derby.

Any type of Dekatron or digital number tubes can be used, some of which are available through Henry's Radio, etc. or Z \& I Aero Services, 44 Westbourne Grove, London, W. 2.

Radiospares do not supply direct to private individuals. Their full range of components is however available to any bona fide retailer or distributor.

## "Most advanced" is now " phased out"

Sir-The receipt of my May 1971 copy of Practical Electronics. containing the editorial entitled "Specialised Components" coincided with a letter from ITT Components Group, Europe, regarding the firm Electroniques of Harlow, Essex.

1 had written to Electroniques regarding certain components specified for the "Transistor D.C. Multimeter", appearing in the June, July and August 1970 issues. Presumably, the supply source of the components in question has been quoted since these were considered "specialised components":

The reply, obviously duplicated. stated simply that the Electroniques operation was being phased out and the components were obsolete. In view of this; the sales depaftment of ITT at Harlow regretted that they were unable to assist.

At the bottom of the letter was the cryptic advertisement, "Europe's most advanced distributor, ITT Electronics Services".

This, I feel, is a rather unfortunate example of, to quote, the editorial: "Industrial component distributors by and large are not favourably disposed towards dealings with private individuals".

In view of this experience I wonder if you could help me with the names of distributors from whom I could obtain the following requirements for the project (not all stated as having been available from the now defunct Electroniques): Miniature Mains Transformer. Primary 0-220/240V, Secondary $12-0-12 \mathrm{~V} \quad 40 \mathrm{~mA}$ (Electroniques
type P9005). 1 mA f.s.d. B.P.L. type S40.V1 moving coil meter. 6 in $x$ 4 in $\times \frac{1}{8}$ in Bakelite, 6 in $\times$ Sin $\times$ ${ }^{\prime}$ I; in Bakelite.
I hope that you are interested in my comments and are able to help with the components.
S. R. Fisher. Ongar, Essex.
We have had no official notification from Electroniques concerning their intention to cease trading. This information has come to us via readers! Some early warning from this company of their intention to "phase out"' the supply of components to private individuals would have enabled us to obviate these current problems, by ceasing to mention Electroniques as a source of supply for various items.

Transformer available from G. W. Smith \& Co (Radio) Ltd.-Eagle type MTI2 240 V primary, $12-0-12 \mathrm{~V}, 50 \mathrm{~mA}$ secondary.

The meter was originally available from Electroniques, but now the SEW type MR-65p from G. W. Smith \& Co will have to be used.

The s.r.b.p. panels should be available from most component stockists.

## WANTED:

A reward is offered for information leading to the arrest of Eddy Current, charged with the induction of an 18 -year-old coil named Milli Henry found induced, half choked and robbed of caluable joules.

This unrectified criminal armed with a ferrite rod, escaped from Weston Primary Cell, where he had been clupped in ions since Faraday.

With an erg to be free, his escape was carefully planned in three phases. First, he fused the electrolytes, then he climbed through a grid despite the impedance of the warders whose reactance was too slow. Finally, he went to earth in a magnetic field.

What seems most likely is that he stole an a.c. motor. This is of low capacity, and he is expected to try to change it for a megacycle, and return by a short circuit to ohm. He may offer series resistance and is a potential killer.
A. C. Maynes- Humm,

Sheriff.


# ASTRONOMY TEERMOUE <br> BY F.W.HYDE• PARTE 

THE simplest radio telescope consisting of aerial, receiver, and recorder is known as a "total power" instrument. It is so named because all the energy received by the aerial is continuously recorded. If the aerial is fixed the result on the recorder will be a "drift curve". That is to say the rotation of the earth will carry the aerial across the sky and record the successive changes in level.

If the aerial is, say, a simple full wave dipole with a reflector pointed up into the sky at an angle of 45 degrees then the two powerful sources of radiation Cygnus A and Cassiopeia will be recorded as humps against the normal background of galactic radiation.
(Cygnus $A$ is part of a nebula sometimes called the Veil Nebula. It is an almost circular streamer of gas about 3 degrees across, and has a high level of radiation extending over a range of frequencies from 100 MHz to $1,200 \mathrm{MHz}$. Cassiopeia $A$ is also the remnants of a nebulae and was discovered by $F$. Graham-Smith. It is thought that these are the remnants of a super nova, parts of it are stationary and parts moving at very high speeds. From calculations of the small moving parts of cloud it would suggest that it actually was a star which exploded about 1702.)

The background galactic radiation shows a gradual rise and fall over the normal 24 hour period. It was this particular variation and its
diurnal change that led Jansky, the pioneer in this branch of astronomy, to make this statement ". . . in conclusion, data have been presented which show the existence of electromagnetic waves in the earth's atmosphere which apparently come from a direction fixed in space. The data obtained give for the coordinates of this direction a right ascension of 18 hours and a declination of 10 degrees."

## A TYPICAL RECORDING

A recording made with the simple, fixed aerial described above connected to a pre-amplifier with a gain of 16 dB , a CR 100 communications receiver and a Recoord pen recorder is shown in Fig. 2.1. Because it was recorded over a period too long to permit the whole recording to be displayed it has been con-certina-ed in sections. The rise of the background can be clearly seen together with the high power sources as the earth sweeps the beam of the aerial across the sky.

With such a simple low power system only the very few really powerful sources can be recorded and even then there may be smaller sources also making up the level but which cannot be resolved. The sun will show up very well and an example is shown in Fig. 2.2.

The frequency at which these records were made was 27.5 MHz .
If the collecting area of the aerial was increased, then there would be an increase in signal level, a narrowing of the aerial beam, and therefore an improvement in resolution. However, this is not the method usually adopted nowadays unless there is a special reason for large areas of collection. Better resolution can be achieved by other means, such as interferometry, which will be described later.

## COMPARISON METHOD

One method adopted to overcome the inherent receiver noise, and so improve the sensitivity, is to

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compare the incoming signal with a standard level of noise produced electronically by a "noise source."

The earliest form of receiver employed for the comparison method was due to Dicke and is known generally as the Dicke system. The receiver is continuously switched between the aerial system and a "noise source". The latter can be a diode valve or semiconductor. The important requirement is the stability of the noise source, since it is the standard or comparison source with which the aerial output is compared.

The Dicke system is shown in the block diagram Fig. 2.3. In (a) is shown the detected levels of $T_{r}+$ $T_{a}$ (receiver noise plus the aerial noise) and $T_{r}+T_{o}$ (receiver noise plus standard source noise) after the detector and before amplification. In (b) is shown the amplitude after the signals passed through the amplifier which is tuned to the same frequency as that of the switching unit. At this point it is an alternating signal. To bring it to a standard level the output of the amplifier is also switched at the same frequency as the switch. The synchronous switching then brings the two signals into phase and thus the two levels show as a varying signal because the standard level is constant but the input from the aerial varies, as shown in (c).

The final trace on the chart will not of course show the divisions as in the diagram (c) which have been exaggerated to make the method clear. The
frequency of the switch is high enough so that a regular change of level appears on the slowly moving pen recorder. The signal level may vary about the level of the standard noise. If the switch frequency is $1,000 \mathrm{~Hz}$ then the intervals shown in the diagram will represent $1 / 1,000$ of a second.

In this system the aerial is connected to the receiver for only half the period and this is an obvious disadvantage. On account of this, such a system is not widely used.

A modification to the Dicke system devised by Graham helps to restore the lost time by using a double receiver. A block diagram for this system is shown in Fig. 2.4.

## RYLE AND VONBERG MODIFICATION

A further modification of the comparison receiver is that due to Ryle and Vonberg. This is based on the Dicke principle of using a noise source for calibration.

In this modification the receiver is switched between the aerial and the calibrating noise source in such a way that the difference signal which appears in the output of the receiver is amplified and controls the noise source so that it is always equal to the input to the receiver. Since this is a varying signal this variation can be recorded. The changing level of the radiation that the aerial collects will thus


Fig. 2.I. A recording of galactic background made with a simple radio telescope
 incoming signals with a standard noise source


Fig. 2.4. The Graham receiving system, which employs a double receiver

FIg. 2.7. A simple two-aerial interferometer


Fig. 2.5. Ryle and Vonberg modification of Dicke system


Fig. 2.6. Cliff interferometer, the sed provides a mirror image of the real aerial
be followed by the recorder. This system also reduces the problems of stability particularly in the receiver. A block diagram of this system is shown in Fig. 2.5.

## LARGE DISH AERIALS

The large parabolic "dish" aerials have enormous collecting areas, and at high frequencies can achieve a very high degree of resolution. For example, the 250 ft dish at Jodrell Bank when used at a frequency of $1,425 \mathrm{MHz}$ has a beamwidth of about 15 seconds of arc, so that two sources at a separation of, say, 25 seconds of arc would be easily distinguished. However, at a frequency of 300 MHz , two sources would have to be more than 45 minutes of are apart before they could be resolved.

This emphasises what has been said previously about the type of aerial system used: it depends essentially upon the nature of the project.

## INTERFEROMETERS

High resolution and increased sensitivity can be obtained by using aerials as interferometers. There are two principal methods used: one is similar in operation to that of the Lloyd's mirror interferometer used in optical work, and the second is equivalent to the Michelson optical interferometer.

The Lloyd's mirror equivalent was devised by McCready, Payne-Scott and Pawsey of Australia. In this arrangement the sea took the place of the mirror for the aerial system was mounted on a cliff overlooking the sea. Thus a single aerial system was all that was necessary for the application of this ingenious experiment. It does, of course, have limitations in that it is suitable for use with sources rising above the horizon to the zenith only.

The cliff interferometer system is shown in the diagram Fig. 2.6. The wave front is intercepted by the sea at the same time that the aerial receives a direct wave. The intercepted wave is reflected from the sea to the aerial but by reason of the longer path of travel there is a difference in phase or time of arrival. The aerial therefore will be subject to additions or subtractions of signal level according to variations of the phase. This means that the receiver will measure a signal increasing or decreasing in power as the changes of phase occur with rising of the source above the horizon and through the beam of the aerial.

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SHURE M30
SHURE M30
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SHURE M44/5/7iC
SHURE M55E
SHURE M55E
SHURE M75E
SHURE M75E
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Fig. 2.8 (a). Diagram of simple interferometer fringes

(b)

## (b). Actual recording from simpie interferometer

If the number of fringes is very large and consequently their width small, there may be some difficulty in correctly identifying the fringe in which the source of radiation appears. In the simple interferometer a diffused source may obscure a smaller one in the same area and a special technique is used to overcome this. The modification is known as phase switching and will be described later.

## LIMITATIONS

One disadvantage of the two aerial interferometer is that the resolution that can be achieved in the north-south direction is not good except when large dishes are used. A dish has the same beamwidth in each mode but the limitations of size imposed by other types of aerial precludes the reduction of the vertical beamwidth to that of the horizontal beamwidth.

For example: at a wavelength of 2 metres four full wavelength dipoles will have a beamwidth of 30 degrees if they are stacked vertically at half-wave spacing. If two of these aerials are spaced 20 wavelengths apart the centre fringe lobe will be a little less than 3 degrees. The area "seen" by such a simple interferometer will be 3 degrees in azimuth and 30 degrees in altitude. Now 40 degrees of sky is wider than the Milky Way which contains a number of sources and it would be difficult to separate them. However, if a study of the sun is to be the object of a project, this system will provide a good deal of data.

## A PRACTICAL RADIO TELESCOPE

At this point it is perhaps appropriate to outline a design which amateurs could put into operation. Three frequencies will be chosen in order that standard equipment can be used. Also the frequencies chosen will enable those with minimum space to at least "have a go"

The three frequencies are $184 \mathrm{MHz}, 144 \mathrm{MHz}$, and 30 MHz . Choice can be made to suit individual needs and preferences and, of course, space limitations. For example if might be easier to raise the frequency of 184 MHz since this is channel 7 in band III and converters are on the market with 19 MHz outputs which may be coupled to a communications receiver, or alternatively, the television receiver can be used without interfering with its normal performance.

At the lower end, 30 MHz is within the communication receiver limit but the physical limit for the
aerial baseline might defeat the aims of the project. At this frequency less than five wavelengths spacing will give rather poor results and might dampen enthusiasm.

The middle frequency 144 MHz should appeal to many, for this is standard for communication and converters for this band are available in quite cheap kit form.

## BASIC REQUIREMENTS

The basic requirements, as we have already seen, are the aerials, a pre-amplifier of 20 dB gain or more, a receiver which may be converter-plus-communication receiver, and (preferably) a pen recorder. As an alternative to the pen recorder, a tape recorder operating at a low speed ( $1 \frac{7}{8} \mathrm{in} / \mathrm{sec}$ or preferably $1 \frac{1}{\mathrm{i}} \mathrm{in} / \mathrm{sec}$ ) could be used for processing to paper later. These basic requirements are indicated in Fig. 2.9.

## AERIALS

The aerial system allows choice of Yagi's, corner reflectors. or pairs of dipoles, or mattress with dipoles (Kooman array). All of these were illustrated in Part 1 of this series.

The aerials must be aligned east and west. The distance between aerials should be correct to one sixteenth of a wavelength. Also remember that magnetic east and west differs by several degrees from geographic east and west. The following table gives the spacings for each frequency.

TABLE 2.1

| $\lambda$ | 184 MHz | 144 MHz | 30 MHz |
| :---: | ---: | ---: | :---: |
| $5 \lambda$ | 21 ft | 30 ft | 150 ft |
| $10 \lambda$ | 42 ft | 60 ft | 300 ft |
| $20 \lambda$ | 84 ft | 120 ft | 600 ft |

## FEEDERS

The aerial feeders should be of identical length. They can be brought individually direct to the receiver, or they may be connected together by a length of cable tapped exactly in the centre from the feeder to the receiver.

In the case of the higher frequencies the preamplifiers should be located at the aerials if possible and they must be fed from stable supplies. If the distance is not too great (say, not more than the distance between aerials) from the aerials to the receiver, then the pre-amplifier can be at the receiver.

At the lower frequency the cable losses will be much smaller so that it will be quite satisfactory if the pre-amplifier is next to the receiver. If the feeder is coaxial cable with an impedance of 75 ohms, then the two cables will need to be matched, either at the junction between the aerials or at the receiver input. A suitable system is shown in Fig. 2.10. The matching section is a length of 50 ohm coaxial, a quarterwavelength long at the selected frequency.

## PEN RECORDER

Pen recorders vary considerably in type and performance specification. Some are fitted with a rectifier so that both a.c. and d.c. operation is possible. If the input is low impedance the recorder can be connected directly to the output of the audio section of the receiver in place of the loudspeaker.


Fig. 2.9. Block diagram of a practical interferometer radio telescope


Fig. 2.10. Matching arrangements for aerial coaxial feeders

If the pen recorder is of the potentiometer type then it may be necessary to connect to some stage prior to the output of the receiver. A suitable point could be the a.g.c. line if the recorder has a d.c. input. If it is of the low resistance a.c. direct reading type then an instrument rectifier will need to be fitted.

## PROJECTS TO BE DESCRIBED

There are a number of radio astronomy projects which can be taken up by individuals or groups working in radio and electronics clubs and schools. Some suitable projects will be introduced in subsequent articles in this series. They will include a study of solar radiation, and radiation from the Galaxy with special-attention to the sources of high level such as Cassiopeia, Cygnus A and the Crab Nebula. It will also be possible to use the various arrangements of apparatus to monitor earth satellites.

Following this there witl be full details of a project for the study of the radiation from the planet Jupiter. Finally, a special project involving a search for evidence of the effects of gravitational waves.
Next month: Increasing the resolving power and
some special methods of observation

## Electronic Microscope

S
hown at the recent Labex exhibition at Earls Court was a new low-cost high-performance electron microscope. The Corinth 275 by GEC-Elliott Automation was shown as the ideal teaching and demonstration instrument with special appeal to biologists requiring a high throughput of routine specimens. The operation of Corinth 275 is very simple with all controls, including a single push-button for the automatic vacuum system, readily to hand. The electron lens system, the vacuum system and all power supplies are self-contained within the single console, requiring a minimum of floor space. The resolution is better than 9 A . The photograph shows the' complete unit.


$\mathrm{A}^{\mathrm{N}}$ ultra high vacuum evaporator manufactured by保 in Europe, was recently installed and is now in production use at ITT Semiconductors, Footscray, Kent.
The unit has a special oil free vacuum system to ensure complete absence of the hydrocarbon contamination present in conventionally pumped equipment.

The evaporator is fitted with a 10 kW electron beam evaporation source and rotary work holder designed for the deposition of aluminium films on $1+\frac{\mathrm{in}}{}$ diameter silicone slices.
During acceptance trials film thickness uniformity and film reliability were both extremely high.

## ELEGTRONQRAMA

Electronic Jay

Opus has been a familiar name to builders of racing engines and cars for a number of years and has now created further interest with the recent announcement of the V12 Jaguar The application of the Opus system on the Jaguar-the first time an electronic ignition system has been fitted as standard equipment to a volume production car.

Lucas Opus Ignition (Mark 2) is an electronic contact breakerless system which will provide a

spark rate of the order of 100 sparks/second. The conventional distributor is replaced by a timing rotor assembly mounted on the drive shaft (see photograph) and a stationary passive pick-up module connected to the amplifier unit. The pick-up module is in effect a transformer having two input windings connected in series and an output winding. When ferrite coupling rods embedded in the timing rotor pass the pick-up the balance of the transformer is upset and increased signal from the oscillator-part of the amplifier module-appears at the output coil. This voltage is fed to the amplifier and causes the output transistor to cut off; the ignition coil primary is thus broken and a spark is fanned by the coil in the normal way. The photograph shows:

1 Distributor h.t. cover
2 Distributor
3 High performance ignition coil
4 Amplifier unit
5 Ballast resistor unit

## I.E.E. CENTENARY

ASPECIAL exhibition to commemorate the centenary of the Institution of Electrical Engineers is now open at the Science Museum, London, S.W.7.
The theme of the exhibition is the development of electrical engineering over the last hundred years.
The exhibition will last approximately another two months and museum times are; 10 a.m. to 6 p.m. weekdays, 2.30 to 6 p.m. Sundays.

## RST

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 N 23 | 0.80 | AC128 | 0.25 | BF181 0.28 | GJ5M | 0-25 | $0 \mathrm{C4} 4$ | $0 \cdot 18$ |
| 1N85 | 0.88 | AC187 | 0.80 | BF184 0-25 | GJ7M | $0 \cdot 38$ | OC44M | 0.18 |
| 1 N 2053 | 0.50 | $\mathrm{ACl}^{\text {c }} 8$ | 0.80 | BF18s 0.25 | H(1000 | 0.50 | $0 \mathrm{C4} 5$ | 0.15 |
| 1N250 | 0.60 | ACY17 | $0-80$ | BF194 0.18 | H8100A | 0.20 | OC45M | 0.18 |
| 1N640 | 0.25 | ACY18 | 0.25 | BF195 0-16 | MAT100 | 0.26 | 0 C 46 | 0.28 |
| 1N72JA | 0.80 | ACY19 | 0.25 | BF196 0.28 | MAT101 | 0.80 | OCST | 0.60 |
| 1 N 914 | 0.08 | ACY20 | 0.28 | BF197 0-28 | Matico | 0.25 | OCJ8 | 0.60 |
| 1N4007 | 0.28 | ACY21 | 0.28 | HFg61 0.28 | MAT121 | 0,80 | OC39 | 0.65 |
| 18021 | 0.20 | ACY22 | 0.18 | BFS98 0.28 | MJEsio | 0.88 | OC66 | 0.50 |
| 18113 | 0.15 | ACY27 | 0.25 | BFX12 0.28 | MJE29JJ | 1.75 | OCro | 0.18 |
| 18130 | $0 \cdot 18$ | ACY28 | 0.18 | BFX13 0.28 | MJ E30ū | 0.98 | 0071 | 0.16 |
| 18131 | 0.18 | ACY39 | 0.55 | HFX29 0.80 | NKT128 | 0.80 | 007: | 0.25 |
| 18202 | 0.88 | ACY40 | 0.16 | BFX30 0.38 | NKT129 | 0.80 | OC73 | 0.80 |
| 2 G 240 | 1.98 | ACY41 | 0.25 | BFX3j 0.88 | NKT211 | 0.25 | 0074 | 0.80 |
| 2G301 | $0 \cdot 18$ | ACY44 | 0.38 | BFX63 0.50 | NKT213 | 0.25 | 0 C 75 | 0.25 |
| 2(3302 | 0-28 | AD140 | 0.80 | $\begin{array}{ll}\text { BFX } 84 & 0.80\end{array}$ | NKT214 | 0.16 | $0 \mathrm{C76}$ | 0.25 |
| -G306 | 0.80 | AD149 | 0.50 | $\begin{array}{ll}\text { BFX83 } & 0.40\end{array}$ | NKT216 | 0.88 | $0 \mathrm{Cr7}$ | 0.40 |
| 2G371 | 0.88 | AD161 | 0.38 | BFX86 0.38 | NKT217 | 0.40 | 0 Cr 8 | 0.20 |
| $2 \mathrm{G381}$ | 0.25 | AD162 | 0.89 | $\begin{array}{ll}\text { BFX } 87 & 0.88\end{array}$ | NKT218 | 0.40 | 0 C 79 | 0.28 |
| $2 \mathrm{G414}$ | 0-30 | AF106 | 0.80 | BFX88 0.20 | NKT210 | 0.88 | $0 \mathrm{OC81}$ | 0.25 |
| 29417 | 0.28 | AF114 | 0.88 | BFY $10 \quad 1.00$ | NKT222 | 0.80 | OC81D | 0.20 |
| 2N214 | 0.48 | AFilij | 0.80 | BFY11 1.25 | NKT224 | 0.23 | OC81M | 0.20 |
| 2 N 247 | 0-25 | AF116 | 0.88 | $\begin{array}{ll}\text { BFY } 17 & 0.25\end{array}$ | NKT251 | 0.24 | OC81DM | 0.18 |
| 2N250 | 0.50 | AF117 | 0.25 | BYF18 0.25 | NKT271 | 0.25 | 0c812 | 0.55 |
| 2N404 | 0.28 | AF118 | 0.68 | BFY19 0.25 | NKT27: | 0.25 | $0 \mathrm{C8} 2$ | 0.25 |
| 2N697 | 0.18 | AF119 | 0.20 | BFY24 0.45 | NKT273 | 0.80 | OC821 | 0.15 |
| 2N608 | 0.48 | AF124 | 0.25 | BFY44 1.00 | NKT274 | 0.20 | 0 C 83 | 0.25 |
| 2N706 | 0.10 | AF125 | 0.20 | BFY 50 0.28 | NKT270 | 0.25 | $0 \mathrm{C8} 4$ | 0.25 |
| 2N706A | 0.18 | AF126 | 0.18 | $\begin{array}{ll}\text { BFY } & 0.20\end{array}$ | NK T277 | 0.80 | OC114 | 0.88 |
| 2N708 | 0.16 | AF127 | 0.18 | BFY52 0. ${ }^{8}$ | NKT278 | 0.25 | OC122 | 0.50 |
| 2N709 | 0.68 | AF139 | 0.80 | BFY  | NKT301 | 0.80 | $0 \mathrm{Cl23}$ | 0.50 |
| 2N711 | 0.38 | AF178 | 0.48 | BFY64 0.48 | NKT304 | 0.86 | OC139 | 0.25 |
| 2 N 987 | 0.68 | AF179 | 0.48 | BFY90 0.68 | NKT403 | 0.76 | OC140 | 0.88 |
| 2N1090 | 0.80 | AF180 | 0.53 | B8X27 0.50 | NKT404 | 0.88 | OC141 | 0.68 |
| 2N1091 | 0.83 | AF181 | 0.48 | B8X60 0.93 | NKT678 | 0.80 | OC169 | $0 \cdot 20$ |
| $2 \mathrm{Nl131}$ | 0.80 | AF186 | 0.40 | B8X 760.16 | NKT713 | 0.96 | 0 Cl 70 | 0.25 |
| 2N1132 | 0.80 | AFY19 | 1.18 | BSY 2600 | NKT773 | 0.25 | 0 Cl 11 | 0.80 |
| 2N1302 | 0.20 | AFZ11 | 0.68 | B8Y27 0.20 | NKT777 | 0.88 | OC200 | 0.88 |
| 2 N 1303 | 0.23 | AFZ12 | 0.75 | B8Y51 0.60 | 0781 | 0.38 | OC201 | 0.48 |
| 2 N 1304 | 0.25 | AsYe6 | 0.25 | B8Y90, 0.15 | OAJ | 0.15 | OC202 | 0.68 |
| 2 N 1305 | 0.25 | ASY27 | 0.38 | B8Y95 0.16 | OA6 | 0.18 | $\mathrm{OCL}^{203}$ | 0.38 |
| 2N1306 | 0.25 | AsY28 | 0.25 | BT102/500R | OA4 4 | 0.10 | OC204 | 0.40 |
| 2N1307 | 0.25 | AsYe9 | 0.80 | 0.76 | OA70 | 0.10 | OC20u | 0.68 |
| 2N1308 | 0.80 | A8Y36 | 0.25 | BTY42 0.98 | OA71 | 0.10 | OC206 | 0.75 |
| 2N1309 | 0.25 | ASY 0 | 0.18 | BTY79/100R | 0.873 | 0.10 | OC207 | 0.75 |
| 2N1420 | 0.98 | $\mathrm{ASY}^{\text {¢ }}$ | 0.40 | 0.75 | 0.174 | 0.10 | OC460 | $0 \cdot 20$ |
| 2N1607 | 0.88 | ARYJ3 | 0.20 | BTY\%9/400R | OA79 | 0.10 | OC470 | 0.80 |
| 2N10326 | 0.38 | ABYos | 0.20 | 1.75 | OA81 | 0.10 | $00^{0} 71$ | 0.98 |
| 2N1909 | 2.25 | ASY62 | 0.25 | BY100 0.18 | OA85 | 0.18 | ORP1: | 0.60 |
| 2N2147 | 0.76 | A8Y86 | 0.88 | BY 1260.15 | 0A86 | 0.16 | ORP60 | 0.40 |
| 2 N 2148 | 0.60 | A8Z21 | 0.48 | BY 1270 | 0 A 90 | 0.10 | ORP61 | 0.48 |
| 2N2160 | 0.68 | AsZ23 | 0.75 | BY127 0.20 | OA91 | 0.08 | S19T | 0.80 |
| 2 N 2218 | 0.80 | AUY10 | 0.88 | BY182 0.76 | OA95 | 0.08 | 8AC40 | O.25 |
| 2 N 2219 | 0.88 | AU101 | 1.50 | BY213 0.25 | OA200 | 0.08 | SFT308 | 0.88 |
| 2N2287 | 1.08 | BC107 | 0.18 | $\begin{array}{ll}\text { BYZ10 } & 0.40\end{array}$ | OA202 | 0.10 | ST722 | 0.38 |
| 2 N 2297 | 0.80 | BC108 | 0.18 | BYZ11 0.85 | 0.4210 | 0.25 | $8 \mathrm{S7231}$ | 0.68 |
| 2N2369A | 0.80 | BC109 | 0.18 | BYZ12 0.80 | OA211 | 0.88 | $8 \times 68$ | 0.20 |
| 2N2613 | 0.88 | BC113 | 0.25 | BYZ12 0.80 | OAZ200 | 0.55 | 8X631 | 0.80 |
| 2 N 2646 | 0.65 | BC110 | 0.88 | BYZ13 0.25 | OAZ201 | 0.50 | 8×63J | 0.80 |
| 2N2712 | 0.25 | BC116 | 0.40 | BYZLis 1.00 | OAZ202 | 0.48 | BX640 | 0.85 |
| 2N2784 | 0.50 | BCl16A | 0.45 | $\begin{array}{ll}\text { BYZ16 } & 0.88\end{array}$ | OAZ203 | 0.48 | 9X641 | 0.85 |
| 2N2846 | 2.25 | BC118 | 0.88 | BYZ88C3V3 | OAZ204 | 0.48 | SX642 | 0.38 |
| 2N2848 | 0.48 | BC121 | 0.80 | 0.18 | OAZ205 | 0.48 | 8X644 | 0.48 |
| 2N2904 | 0.80 | BC122 | 0.20 | C111 0.65 | OAZ206 | 0.48 | 8X640 | 0.75 |
| 2 N 2904 A | 0.88 | BC12 | 0.88 | CRS1/05 0.25 | OAZ207 | 0.48 | V15/30P | 0.60 |
| 2N2906 | 0.40 | BC126 | 0.65 | CRS1/40 00.48 | OAZ208 | 0.88 | V $30 / 201 \mathrm{P}$ | 0.88 |
| 2N2907 | 0.88 | BC140 | 0.55 | $\mathrm{Cs}_{4} \mathrm{~B}$ 8 8.50 | OAZ209 | $0 \cdot 88$ | V60/201 | ${ }_{0}^{0.88}$ |
| 2 N 2924 | 0.88 | BC147 | 0.18 | C810B 8.18 | OAZ210 | 0.88 | V60/201P | 0.38 |
| 2 N 2925 | 0.18 | BCl48 | 0.18 | DD000 0 | OAZ211 | 0.88 | XA101 | 0.10 |
| 2N2926 | 0.18 | BC149 | 0.20 | DD003 0.15 | OAZ222 | 0.40 | XA102 | 0.18 |
| 2N3054 | 0.50 | $\mathrm{BC1}^{\mathrm{BC} 59}$ | 0.80 | $\begin{array}{ll}\text { DD006 } & 0.18\end{array}$ | OAZ223 | 0.40 | XAl51 | 0.15 |
| 2 N 3055 | 0.75 | BC157 | 0.20 | ${ }_{\text {DD } 007} \quad 0.40$ | OAZ224 | 0.88 | XAlu'2 | 0.15 |
| 2N3702 | 0.18 | BC158 | 0.80 | DD008 0.88 | OAZ241 | 0.88 | X Al 161 | 0.25 |
| 2N3705 | 0.15 | BClb0 | 0.68 | GD3 0.88 | OAZ242 | 0.28 | XA162 | 0-25 |
| 2 N 3706 | 0.88 | BC169 | 0.18 | GD4 0.05 | OAZ244 | 0.28 | XA162 | 0.48 |
| $2 N 3707$ | 0.15 | BCY31 | 0.80 | GDJ 0.88 | OAZ248 | 0.88 | XB101 | 0.48 |
| 2N3709 | 0.18 | BCY32 | 0.50 | GD8 0 | OAZ290 | 0.88 | XB102 | 0.10 |
| 2 N 3710 | 0.18 | BCY32 | 0.80 | GD12 0.05 | 0 OCl 6 | 0.48 | X E103 | 0.25 |
| 2N3711 | 0.18 | BCY33 | 0.80 | GET102 0.30 | 0 OCl 6 T | 0.88 | X 8113 | 0.10 |
| 2N3819 | 0.85 | BCY34 | 0.25 | GET103 0.88 | OC19 | 0.88 | XB191 | 0.48 |
| 2N3820 | 0.88 | BCY 98 | 0.80 | GET113 0.20 | $\mathrm{OCl2}^{0}$ | $0 \cdot 08$ | XBlel | 0.48 |
| 2 N 3823 | 0.75 | BCY 39 | 0.48 | GET114 0.15 | OC22 | 0.48 | ZRed | 0.68 |
| 2N0027 | 0.58 | BCY40 | 0.48 | GET110 0-45 | OC83 | 0.50 | 28170 | 0.10 |
| 2N5088 | 0.88 | BCY42 | 0.15 | GET116 0.50 | OC24 | 0.50 | 28271 | 018 |
| 28000 | 0.75 | BCY70 | 0.80 | GET120 0.26 | $00^{2} 25$ | 0.88 | ZT21 | 0.25 . |
| 28301 | 0.48 | BCY71 | 0.80 | GET872 0-80 | OC26 | 0.25 | ZT43 | 0.2.5 |
| 28304 | 0.88 | BCZ10 | 0.80 | GET875 0.85 | OC28 |  |  |  |
| 28001 | 0.88 | BCZ11 | 0.88 | GET880 0.88 | OC28 | 0.68 | ZTX107 | 0.15 |
| 28703 | 0.68 | BD121 | $0 \cdot 65$ | GET881 0.25 | 0 C 29 | 0.68 | ZTX108 | 0.15 |
| AA129 | 0.20 | BD123 | $0 \cdot 88$ | GET882 0.25 | OC30 | 0.40 | ZTX 300 | 0.18 |
| AAZ12 | 0.80 | BD124 | 0.68 | GET885 0-20 | 0 C 35 | 0.50 | ZTX304 | 0.88 |
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| AC107 | 0.88 | BF117 | 0.60 | GEX941 0.15 | $0 \mathrm{OC4} 1$ | 0.25 | ZTX503 | 0.20 |
| AC126 | 0.2\% | BF167 | 0.25 | GJ3M 0.26 | OC42 | 0.80 | ZTX531 | 0.80 |

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# Super IC-12 <br>  

## HighfidelityMonolithic Integrated Circuit Amplifier

Two vears ago Sinclair Radionics announced the World's first monolithic integrated circuit $H_{I}-F_{1}$ amplifier, the IC.10. Now we are delighted to be able to introduce its successor the Super IC. 12. This 22 transistor unit has all the virtues of the original IC. 10 plus the following advantages:

1. Higher power.
2. Fewer external components
3. Lower quiescent consumption.
4. Compatible with Project 60 modules.
5. Specially designed built-in heat sink.

No other heat sink needed.
6 . Full output into $3,4,5$ or 8 ohms.
7. Works on any voltage from 6 to 28 volts without adjustment.
8. NEW 22 transistor circuit.

Output power 6 watts RMS continuous 112 watts peak).
Frequency Response 5 Hz to $100 \mathrm{KHz} \pm$ $\vdots \mathrm{dB}$.
Total Harmonic Distortion Less than $1 \%$ (Typical $0.1 \%$ ) at all output powers and all frequencies in the audio band.
Load Impedance 3 to 15 ohms.
Power Gain 90dB (1,000,000,000 trmes)子fter feedback.
Supply Voltage 6 to 28 volts (Sinclair PZ-5 or PZ-6 power supples ideal).
Size $22 \times 45 \times 28 \mathrm{~mm}$ including pins and neat sink.
Input Impedance 250 Kohms nominal.
Quiescent current 8 mA at 28 volts.
Price: including FREE printed cricult board for mounting. $\mathbf{£ 2 . 9 8}$ Post free

With the addition of only a very few external resistors and capacitors the Super IC. 12 makes a complete high fidelity audio amplifier suitable for use with pick-up. F.M. tuner etc. Alternatively, for more elaborate systems. modules in the Project 60 range such as the Stereo 60 and A.F.U. may be added. The comprehensive manual supplied with each unit gives full circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include car radios, oscillators etc. The very low quiescent consumption makes the Super IC. 12 ideal for battery operation.

[^2]

## Sinclair Project 60



## the world's most advanced high fidelity modules

Sinclair Project 60 presents high fidelity in such a way that it meets every requirement of performance, design. quality and value and now that the remarkable phase lock loop stereo FM tuner is available, it becomes the most versatile of high fidelity systems. With Project 60, it is possible to start with a
modest mono record reproducer and expand it to a sophistrcated stereophonic radio and record reproducing system of fantastically good quality to hold its own with any other equipment, no matter how expensive. Project 60 is a unique high fidelity module system where compactness and ease of assembly are combined with

|  | System | The Units to use | together with | Cost of Units |
| :---: | :---: | :---: | :---: | :---: |
| A | Simple battery record player | 2.30 | Crystal P.U., 12 V battery volume control | £4.48 |
| B | Mains powered record player | Z.30, PZ.5 | Crystal or ceramic P.U. volume control etc. | ¢9.46 |
| C | 20+20W. R.M.S. stereo amplifier for most needs | $\begin{aligned} & 2 \times 2.30 s, \text { Stereo 60, } \\ & \text { PZ.5 } \end{aligned}$ | Crystal, ceramic or mag. P.U.. most dynamic speakers. F.M. tuner eic. | £23.90 |
| D | $20+20$ W. R.M.S. stereo amplifier with high performance spkrs. | $\begin{aligned} & 2 \times 2.30 \mathrm{~s}, \text { Stereo } 60, \\ & \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetic P.U., F.M. Tuner. Tape Deck, etc. | £26.90 |
| E | $40+40 \mathrm{~W}$. R.M.S. delixe stereo amplifier | $2 \times 2.50 \mathrm{~s}$, Stereo 60 PZ.8, mains trsfrmr | Asfor D | £34.88 |
| F | Outdoor P.A. system | 2.50 | Mic., up to 4 P.A. speakers controls. etc. | ¢5.48 |
| G | Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers, etc., controls | $£ 19.43$ |
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|  | Radio | Stereo F. M. Tuner | C. Dor E | £25.00 |

circuitry that is far in advance of any other manufacturer in the world. Thus it is extraordinarily easy to assemble any combination of modules using nothing more complicated than the simplest of tools, and you certainly do not have to be experienced to build with complete confidence. The 48 page manual free with Project 60 equipment makes everything easy and you can house your assemhly in an existıng cabinet, motor plinth, free standing cabinet or virtually any arrangement you wish. Once you have completed your assembly you will have superlatively good equipment to give you years of service and enjoyment. You will have obtained superb value for moneybecause Project 60 is the best selling modular system in Europe and can therefore be produced at extremely competitive prices and with excellent quality control.
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## Sinclair Project 60

## Z. 30 \& Z. 50 power amplifiers



The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at full output and all lower outputs. Whether you use $Z .30$ or $Z, 50$ ampliffers in your Project 60 system will depend on personal preference. but they are the same size and may be used with other units in the Project 60 range equally well.
SPECIFICATIONS (Z50 units are interchangeable with $Z .30$ sin all applications).

## Power Outpurs

Z.30 15 watts R.M.S. into 8 ohms using 35 volts: 20 watts R.M S. ".to 3 ohms using 30 volts.
$\mathbf{Z . 5 0} 40$ watts R MS into 3 ohms using 40 volts: 30 watts R.M S. into 8 ohms. using 50 volts.
Frequency response: 30 to $300000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$.
Distortion: $002 \%$ into 8 ohms.
Signal to noise ratio: better than 70 dB unweighted.
Input sensitivity: 250 mV into 100 Kohms.
For speakers from 3 to 15 ohms impedance.
Size $3 \frac{1}{2} \times 2 \frac{1}{4} \times \frac{1}{2} \mathrm{in}$
2.30

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## Power Supply Units



Designed specially for use with the Project 60 system of your chorce.
Illustration shows PZ. 5 to left and $P Z .8$ (for use with Z.50s) to the right. Use PZ.5 for normal Z.30 assemblies and PZ. 6 where a stablised supply is essential
PZ.5 30 volts unstabilised $£ 4.98$
PZ-6 35 volis stabilised $\mathbf{£ 7 . 9 8}$
PZ-8 45 volts stabillsed
(less mains uansformer) $£ 7.98$
PZ-8 mans transformer $£ 5.98$

## Stereo 60 pre-amp/control unit



Designed for the Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high stgnal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is prowded for all the usual inputs.

## SPECIFICATIONS

Input sensitivities: Radio-up to 3 mV . Mag. p.u. 3 mV . correct to R.I.A.A. curve $\pm 1 \mathrm{~dB}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u.-up to 3 mV : Aux-up to 3 mV .
Output: 250 mV
Signal-to-noise ratio: better than 70 dB .
Channel matching: within 1 dB .
Tone controls: TREBLE +15 to -15 dB at 10 KHz . BASS $+15 \mathrm{to}-15 \mathrm{~dB}$ at 100 Hz
Front panel: brushed aluminium with black knobs and eontrols.
Size: $8 \frac{1}{4} \times 1 \frac{1}{2} \times 4 \mathrm{ins}$
Bult. tested
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## Active Filter Unit <br> 

For use between Stereo 60 unit and two $Z .30$ s or $Z 50$ s. and is easily mounted. It is unique in that the cut-off frequencies are contonuously variable, and as attenuation in the rejected band is rapid ( $12 \mathrm{~dB} / o c t a v e$ ), there $: s$ less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are neghigible. The A.F.U. is suitable for use with any other amplifier system. Two stages of filtering are incorporated rumble (high pass) and scratch (low pass). Supply voltage - 15 to 35 V . Current - 3 mA . H.F cut-off ( -3 dB ) variable from 28 kHz to 5 kHz . L.F cut-off ( -3 dB ) variable from 25 Hz to 100 Hz . Distortion at $1 \mathrm{kHz}(35 \mathrm{~V}$. supply) $0.02 \%$ at rated output.
Bult tested
and guaranteed
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## Stereo FM Tuner



## first in the world to use the

phase lock loop principle
Before production of this tuner. the phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio over other systems. Now for the first time, the principle has been applied to an FM tuner with fantastically good results. Other original features include varicap diode tuning. printed circuit coils. an I.C. in the specially designed stereo decoder and squelch circuit for silent tuning between stations. Sensitivity is such that good reception becomes possible in difficult areas. Foreign stations can be tuned in suitable conditions and often a few inches of wire are enough for an aerial. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with any other high fidelity system.

SPECIFICATIONS:
Number of transistors: 16 plus 20 in I.C.
Tuning range : 87.5 to 108 MHz
Capture ratio: 1.5 dB
Sensitivity: $2 \mu \mathrm{~V}$ for 30 dB quieting: $7 \mu \mathrm{~V}$ for full limiting.
Squalch level: $20 \mu \mathrm{~V}$.
A.F.C. range: $\pm 200 \mathrm{KHz}$

Signal to noise ratio: $>65 \mathrm{~dB}$
Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}$ $( \pm 1 \mathrm{~dB}$ )
Total harmonic distortion: $0.15 \%$ for $30 \%$ imodulation
Stereo decoder operating level: $2 \mu \mathrm{~V}$
Pilot tone suppression: 30 dB
Cross talk: 40 dB
I.F. frequency: 10.7 MHz

Output voltage: $2 \times 150 \mathrm{mVR}$.M.S.
Aerial Impedance: 750 hms
Indicators: Mains on: Stereo on; tuning indicator
Operating voltage: $25-30 \mathrm{VDC}$
Size : $3.6 \times 1.6 \times 8.15$ inches: $91.5 \times 40 \times 207 \mathrm{~mm}$


Price: $\mathbf{5} 25$ built and tested. Post free

## Guarantee

If within 3 months of purchasing project 60 modules directly from us. you are dissalisfied with them. we will refund your money at once. Each module is guaranteed to work pe fectly and should any defect anise in normal use we will service It at once and without any cost to you whatsoever provided that It is returned to us within 2 vears of the purchase date. There will be a small charge for service thereafter No charge for postage by surface mail. Aur-mall charged at cost.

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## Sinclair Q16/Micromatic

## Q16 High fidelity loudspeaker

The Q16 employs the well proven acoustic principles specially developed by Sinclair in which a special driver assembly is meticulously matched to the characteristics of the uniquely designed cabinet. In reviewing this exclusive Sinclair design. technical journals have justly compared the Q16 with much more expensive loudspeakers. Its shape enables the Q19 to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures. A solid teak surround with a special all-over cellular foam front is used as much for appearance as its ability to pass all audio frequencles without loss.

This elegantly designed shelf mounting speaker brings genuine high fidelity within reach of every music lover.

## Specifications:

Construction: Special sealed seamless sound or pressure chamber with internal baffle.
Loading: up to 14 watts RMS
Input Impedance: 8 ohms.
Frequency response: From 60 to 16,000 Hz . confirmed by independently plotted B and K curve.
Driver unit: Special high compliance unit having massive ceramic magnet of 11,000 gauss, aluminium speech coil and special cone suspension for excellent transient response.
Size and styling: $9 \frac{3}{4} \mathrm{in}$. square on face $x$ $4 \frac{3}{4}$ in. deep with neat pedestal base. Black all over cellular foam front with natural solidteak surround.
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## Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage AM receiver brilliantly designed to provide remarkable standards of selectivity, power and quality for its size. Powerful AGC counteracts fading from distant stations: bandspread at higher, frequencies makes reception of Radio 1 easy. The plug-in magnetıc earpiece provided, matches the Micromatıc's output to give wonderful standards of reproduction. Everything including the special ferite rod aerial and batteries is contained within the minute attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built and tested, you will find It as easy to take with you as your wrist watch, and dependable under the severest listening conditions.

Specifications:

Size: $36 \times 33 \times 13 \mathrm{~mm}(1.8 \times 1.3 \times 0.5 \mathrm{in}$.) Weight: including batteries, 28.4 gm (10z.)
Case: Black plastic with anodised alumınium front panel and spun aluminium dial.
Tuning: medium wave band with bandspread at higher frequencies (550 to $1,600 \mathrm{KHz}$ ).
Earpiece: Magnetic type.
On/off switching: By inserting and withdrawing earpiece plug.

Kit in pack with earpiece, case, instructions and solder $\mathbf{£ 2 . 4 8}$

Ready built, tested and guaranteed, with earpiece $\mathbf{£ 2 . 9 8}$.
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MTV3 $20-0-20 \mathrm{~V}, 0.75 \mathrm{~A}, 81-20$ each
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 $8 \times 5 \mathrm{in}, 81.35,10 \times 6 \mathrm{in}, 81 \cdot 90$.
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| 600 | 0.53 | 0.57 | 0.77 | 0．97 | 1.25 |  |
| 00 | 0.63 | 0.70 | 0.80 | 1.20 | 1.50 | 4.00 |

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{Ep}^{\text {p }}$ | fp | fp | £p | fp | £ | ep |
| 50 | 0.04 | 0.05 | 0.05 | 0.07 | 0－14 | 0.21 | 0.47 |
| 100 | 0.04 | 0.08 | 0.05 | 0.13 | 0.16 | 0.23 | 0.75 |
| 200 | 0.05 | 0.09 | 0.68 | 0.14 | 0.20 | 0.84 | 1.00 |
| 400 | 0.06 | 0.13 | 0.07 | 0.20 | 0.27 | 0.37 | 1.25 |
| 600 | 0.07 | 0.16 | 0.10 | 0.23 | 0.34 | 0.45 | 1.85 |
| 800 | $0 \cdot 10$ | 0.17 | 0.13 | 0.25 | 0.37 | 0.55 | 2.00 |
| 1000 | 0.11 | 0.25 | 0.15 | 0.30 | 0－46 | $0 \cdot 63$ | 2.50 |
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| VBOM 2 A |  | 1.1 | 10.1 |  |  |  |  |
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|  | 25 | 2p | 2p | Fhying Leadx，80p each． |  |  |  |
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C14 150 Mixed silicot and germanimm liosles

$103 \cdot A$ inp sificon rect thersm mad type up to 1000 IIS
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$$
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\begin{aligned}
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