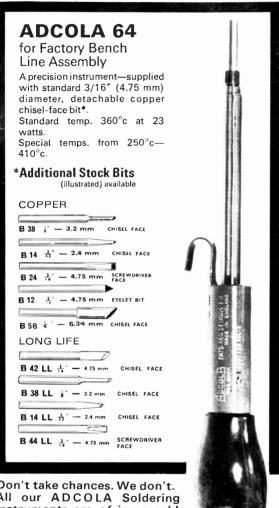
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Huge 6in Meter Scale ⊕ Ranges: d.c. n-0-5-15-50-150-5001 full scale. ♠ Accuracy: ±1-3% of full scale reading funput Resistance: 1/M Ω to 1.5001 ⊕ Ranges: a.c. n-1-5-5-15-50-150-500-1,5001 full scale. n-2-11-42-140-120-1,5001 full scale. n-2-11-42-140-120-1,5001 full scale. n-2-11-42-140-120-1,5001 full scale. n-2-11-42-10-1,000 full scale. n-2-11-40-10-1,000 ohms: n-1-10-100-1,000 full scale. n-1,000-100-1,000 ohms: n-1-10-100-1,000 full scale. n-1,000-12-13 € CSS. KG 620 as KG 625 (Same Ranges) but 4½ in Scale: 9½ GMS.



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Designed to Satisfy the Most Critical Listener



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R.F. and I.F. Fanels Assembled and Aligned Front Panel Stereo Headphone Jack and Speaker Muting Switch Fall Controls including Tape Monitor 550W I.H.F. Output 0 Usable I.H.F. Sensitivity: 3\(\mu^{1}\) = Frequency Response: \(\plu^{1}\) IdB. 30 to 1\(\plu^{1}\) \(\precedef{Armonic Distortion: less than 10\) \(\phi^{0}\) A.M. Suppression: more than 30d B \(\phi\) Complementary

Transformerless \(\mathbf{Power}\) Power Bandwidth: 20-20 \(\pi\) \(\pi\) undits at 1\) ker according Response: \(\pi\) 1dB. 1s to 30\(\phi\) \(\pi\) Hills \(\phi\) Power Bandwidth: 20-20 \(\pi\) \(\pi\) \(\pi\) at monitor distortion: \(\mu\) Expansion: \(\pi\) Two R.F. Suppression: \(\pi\) and \(\pi\) \(\pi\)

STAR ROAMER 5-Band Shortwave Receiver Kit

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KG-640 20,000 ohms/Volt VOM Kit

BIG PROFESSIONAL SIZE 62 RANGE MULTIMETER

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Scale accuracy: DCV and current: ±2%, ACV: ±3%, circult measurements.

Special of transistor circult measurements.

SPECIFICATION

SPECIFICATION

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Stereo Magnetic Cartridge. Frequency response: 20-20,000Hz. Output: SmV. Stylus: Diamond LP. Post Free Tracking force: 2 gms ±0-5 gm. Replacement stylus type US.P1 41/- post free.

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1-0-1mA 2mA 5mA 10mA 20mA 50mA 100mA

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50mA

100m 4

500mA

ImA

TYPE S-80



5 1 d e

150mA 200mA

300m \

500m A

750mA

2.1...

10.5

3V d.

10V d.c. 15V d.c. 20V d.c.

100V d.c. 150V d.c. 300V d.c.

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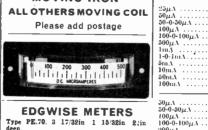
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2N1306 2N1307	5/- ASY28 5/- ASZ17 5/- AUY10 5/- BAY31 6/- BAY31	19/6 NKT104: 1/6 OA5	89 7/6 8/-	6BR*
9N1308	6/- BC107 5/- BC108		5/-	6BW6 6BW7
2N 1309 2N 1613	4/6 DC100	2/6 OA47 2/6 OA70 5/- OA79	2/- 2/-	6BZ6 6C4 6CD6
2N 1711	9/* BC113	5/- OA79 OA81	2/-	6CD6
2N1889 2N1893	10/- BC125	8/- OA81 11/- OA85 11/- OA90	2/6	6CL6 6CW4
2N2147 2N2160 2N2193	15/- BC126 12/6 BC147	11/- OA90 3/6 OA91	2/- 1/6 1/6	6F1
2N2193	9/6 BC148	2/6 OA95	1/6	6F6G 6F13 6F14
2N2217 2N2218	5/6 BC149 6/- BC167	4/- OA202	1/6 2/-	6F14 6F15
2N2219	6/6 BC172 8/6 BC177		3/6 7/6	6F18 6F23
?N2369 2N2369	4/- BC186	7/6 OC19 7/6 OC20	19/6	6H6
2N2369A 2N2484		4/- 0022	9/6 10/-	6J4 6J5
2N2613 2N2646	7/- BC184L BC212L 10/6 BCY30	4/6 0024	10/- 7/6	6J5(:T
2N2904	6/- BCY30 8/6 BCY32 8/6 BCY32 8/6 BCY33	5/- OC26 6/- OC28	5/-	6J6 6J7
2N2923 2N2924	8/6 BCY32	10/- OC28 OC29	12/6 12/6	6K8G 6L6GT
2N2925 2N2926G	3/6 BCY33 3/6 BCY34 2/6 BCY38 2/6 BCY42 2/6 BCY43 5/- BCY72 10/- BCY72		10/- 12/6	6LD20 6Q7
2N29260 2N2926Y	2/6 BCY42	8/. OC41	5/-	14,60
2N3053	5/- BCY43 BCY71	6/- OC42	6/- 8/6	68G7 68J7
2N3054	10/- BCY72 15/- BCZ11	a (alDC45	8/- 5/6	68K7 68L7
2N3055 2N3391A	6/- BD121 7/6 BD123 2/6 BD124 17/6 BF115 2/6 BF117 8/6 BF117	18/- 0070	2/6	68N7
2N3416 2N3570 2N3702	2/6 BD123	16/6 OC72 12/- OC72	8/- 5/-	68Q7 6U4 6V6G
2N3702 2N8703	17/6 BF115	5/- OC73	6/- 6/-	6V6G 6V6GT
2N3704	8/6 BF167	16/6 OC72 18/- OC73 5/- OC74 5/- OC74 5/- OC74	5/- 5/-	6X4
2N3705 2N3706	3/- BF173 4/6 BF180 3/- BF181	6/- OC76 7/6 OC77 7/6 OC78	5/- 8/-	6X5G 6X5GT
2N3707	8/- BF181		4/- 5/-	110C2
2N3708 2N3709 2N3710	3/6 BF181 2/6 BF184 2/6 BF185 2/6 BF194	6/6 OC81 5/- OC811 5/- OC83	4/-	10F1 10P13
2N3710 2N3711	2/6 BF185		5/-	10P14 12AT6
2N3819 2N3903		8/- OC139	5/- 7/6	12AT6 12AT7 12AU7
2N3904	7/- BF200 7/- BF224 7/6 BF225	8/6 OC149 7/6 OC140 6/- OC169	49/m	12AX7 12AY6
2N3905	7/6 BF225	a/ IOC170		12AY6 12BA6
2N3906 2N4058	7/6 BF244 8/6 BFX12 5/- BFX13	9/6 OC171 4/6 OC200 4/6 OC201 6/- OC202	6/- 7/6	12BE6
2N4059 2N4061	5/- BFX13	4/6 OC201	9/6 12/6	19AQ5
2N4062	4/6 BFX 30		7/6	20D1 20F2
2N4286 2N4287	8/6 BFX 85	6/6 OC203 7/6 OC204 8/- OC205 6/6 OC207 6/6 OCP71 5/- ORP12	8/- 12/6	20L1
2N4288 2N4289	3/- BFX86	6/6 OC207	15/- 19/6	20P1 20P3
2N 4290	8/- BFX88	5/- ORP12	10/-	20P4 20P5
2N 4291 2N 4292	8/- BFY18	5/- OKP60 12/6 P346A	8/- 5/- 2/9	
2N5354 2N5355	5/6 BFY50	6/6 OCP71 5/- ORP12 5/- ORP60 12/6 P346A 4/6 PL4001 4/- PL4002	2/9 3/-	400m
28102	5/- BFY52	4/6 PL4003	8/8	1.5 wa
28103 28104	7/6 BFY90 7/6 B8X19	4/6 PL4003 13/6 PL4004 3/6 PL4005 3/6 PL4006	3/6 3/9	. —
40250 40361	5/- B F X 33 3/- B F X 33 3/- B F X 33 3/- B F X 34 3/- B F X 34 3/- B F X 36 3/- B F X 37 3/- B F Y 20 5/6 B F Y 50 5/- B F Y 50 7/8 B F X 90 7/8 B F X 90 11/- B F X 20 11/- B F X 20	8/6 PL4006	4/-	
40362	12/- BSX76 ·	7/6 PL4007 3/- TIS43 8/6 TIS44	8/-	1.000
AC107 AC126	7/6 B8Y26 5/- B8Y27 5/- B8Y28		2/6 3/6	L900 § 12/6;
AC127	5/-1BSY28	3/6/TIS46	8/6	42/8; N 55/-; 1

THYPICTORS

IIIIIIIII							
PIV	50	100	200	300	340	400	
LA	5/-	5/6	7/6	8/-		9/6	-
3A	8/-	7/6	8/-	9/-		10/-	
5.A.		11/-	-18/-	www	_	15/-	
7A		11/-	13/-	14/-	-	19/6	
1004			,		50/-	EE/	- 1

VALVES

2/6 25L6	7/6 EL91 6/- EL95
7/- 25Z4	6/- EL95
5/- 257.5	8/6 EM80
4/6 25Z6	11/6 EM81
8/- 30C15	15/- EM84
9/6 30C17	16/- EM85
6/6 30C1N	15/- EM87
6/6:30F5	17/- EY51
6/- 30FL1	15/- EY86
2/6 30FL12	18/6 EY87
6/6 30FL14	15/6 EZ40
8/- 30L15	17/- EZ41
7/- 30L17	17/- EZ80
8/- 30P12	16/- EZ81
11/- 30P19	15/- GZ32
6/- 30PL1	15/6 GZ34
8/- 30PL13	18/- K166

9/-5/6 5/6 9/6 11/-27/6 88/6 6/- 30PL14

8/-

9/6 7/-8/-12/-10/6 11/-6/6 6/9 9/-

13/-15/6 7/9 13/-8/9 9/6 9/6

15/-16/-11/-12/6

6/6

8/-

10/-15/-15/-6/-6/-

14/-8/-8/-

11/6

20/-

10/6 9/6 8/-7/8

7/-9/9 8/-10/6 13/-

10/-7/6

12/-6/6 8/-

7/6 33Lif 15/- 33W4 4/- 33Z4 7/6 32Z4 7/6 32Z4 7/6 3Z4 7/

3/- 6BR%
5/- 6BW7
2/- 6BW7
2/- 6BBW7
2/- 6BC4
2/- 6C4
2/- 6C4
2/- 6C4
1/- 6C4

6/- EBC81
3/- EBF89
4/- EBC88
6/- ECC49
6/- ECC49
6/- ECC49
6/- ECC49
6/- ECC89
6/- EC 2/6 6Q7 5/- 68A7 6/- 68G7 8/- 68J7 8/- 68L7 2/6 68L7 2/6 68N7 3/- 68Q7 5/- 6V6G

6/- 6V6G 6/- 6V6G 6/- 6V6G 5/- 6V6G 5/- 6X5G 4/- 10F13 5/- 10F13 5/- 10P13 5/- 12AT7 7/6 12AT7 7/6 12AX7 5/- 12AX7 6/- 12BX6 9/6 12BE6 9/6 12BA6

12BE6 12BH7 19AQ5 20D1 20F2 20L1

20P1 20P3 20P4 20P5

ZENER DIODES 400mw. (3.3 to 33v) 3/-1.5 watt (2.4 to 200v) 4/-10 watt (3.9 to 100v) 5/-

INTEGRATED **CIRCUITS**

CRCUITS
L900 9/9; L923
L9:6: TC-10 50/-: NL403
L9:6: CA-2005
L9:6: CA-2005
L9:6: CA-2005
L9:6: CA-2005
L9:6: TC-H131 10/-: FJH-11
L9/-: FJH-11 10/-: FJH-11
L9/-: FJH-11 10/-: FJH-11
L9/-: FJH-12 10/-: FJH-11
L9/-: FJH-12 10/-: FJH-13
L9:6: FJJ-12 17/6: FJJ-13
L7:6: FJJ-12 17/6: FJJ-13
L7:6: FJJ-12 17/6: FJJ-10
L9/-: FJJ-10/-: FJJ-10 10/-

SEND SAE FOR FULL LISTS:

DISCOUNTS 10% on 12+ any one type. 15% on 25+ any one type. Large quantity discounts on application Postage—

Semi Conductors 1/6; Valves 3/-

HI-FI EQUIPMENT SAVE UP TO 33% OR MORE

SEND S.A.E. FOR DISCOUNT PRICE LISTS AND PACKAGE OFFERS!



FULL CURRENT RANGE OFFERED, BRAND NEW AT FANTASTIC SAVINGS

224,19,6 \$16,19,6 \$25,19,6 \$27,19,6 \$27,10,0 \$11,19,6 \$25, 0,0 8L72B AP75 8L75 2025T/C stereo 28.17.6 40 MKII stereo £8,8,0 3000 stereo £9,19,6 SL75B SL95B A70/11 8P25 MK111 411,15,0 SLAS £11.12.6 SL65B £14,19,6 Carriage 7,6 extra each item.



TEAK PLINTHS AND PERSPEX COVERS
1. For SP26, SL56, SL56, 3050, 2025T/C, 2025, 1000. 24.10.0
2. For AF76, SL75, SL95, 25.198, 25.198
3. For SP26 etc. to operate with lid in place 25.19.6
Carriage 7/6 extra each type.
Full range of Garrard accessories available.

TELETON SPECIAL OFFER!



CRIOT AM/FM STEREO TUNER AMPLIFIER WITH MATCHING PAIR SALOOS SPEAKER SYSTEMS

Output 4 watts per channel. Excellent reception AFC, built-in MPN. Cer/NTAL Input. Total List \$50.5.0. OUR PRICE \$29,19.0, Carr. 12/6.

Also available with Garrard 2025T/C Record Changer, Plinth, cover and stereo cartridge. Ready wired. 245. Carr. 20/-.

TELETON SAQ203E STEREO AMPLIFIER



Solld state. 10W per channel music power. 6W r.m.s. Inputs for mag, power. 6W r.m.s. Inputs for mag, ceramic, tuner, aux.

ECHO HS-606 STEREO HEADPHONES



Wonderfully com-Wondermy, fortable. Light-weight adjustable vinyl headband. 6ft. cable and stereo jack plug. 25-17,000 cps. 8 cbm imp. 67/8. ohm imp. P. & P. 2/6.

HOSIDEN DHO4S 2-WAY STEREO HEADSETS



Each headphone contains a 2½n woofer and a 3½n tweeter. Built in individual level controls. 8 Ω imp. 25-18,000c/s. with cable and stereo plug. \$5.19.6. P. & P. 2/6.

AMERICAN TAPE

AMERICAN TALE	А
First grade quality American tapes. Brand new Discount on quantities.	ı
3in. 225ft. L.P. acetate	d
31in. 600ft. T.P. mylar 10/-	
5in. 600ft. std. plastic	
5in. 900ft, L.P. acetate 10/-	
5in. 1.200ft. D.P. mylar 15/-	
58in, 1.200ft, L.P. acetate 12/6	ı
5tin. 1,200ft, L.P. mylar 16/-	
5 in. 1,800 it. D.P. mylar 22/6	
53in. 2,400ft. L.P. mylar	
7in. 1,200ft. std. acetate 12/6	
7in. 1,800ft. L.P. acetate	1
7in. 1,800ft, L.P. mylar 20/-	1
7in. 1,800ft, L.P. mylar	1
7ln. 3,600ft. T.P. mylar 45/-	1

SPECIAL OFFERS!

Garrard SP25/HI fitted Goldring G800 cartridge and wooden plinth and plastic ever. Total list price \$25.

OUR PRICE \$20.19.8, Carr. 19/-.

GOLDRING GL89/S fitted Goldring G800 cartridge complete with de luxe base and cover. Total list price \$20.16.0.

OUR PRICE \$30, Carr. 20/-.

SINCLAIR EQUIPMENT



SINCLAIR EQUIPMENT
Project 60, Package offers. 2 × 230 amplifier, stere 60 pre-amp. P.25 power supply, \$19.0, Carr. 7/6. Or with P26 power supply, \$21, Carr. 7/6. 2 × 250 amplifier, stere 60 pre-amp. P.28 power supply, \$21.10, Carr. 7/6. Transformer for P24. 55% extra. Add to any of the above \$4.17.6. for active filter unit and \$16 for a pair of Q16 speakers. Project 60 Fm Tuner \$20.19.8, Carr. 7/6.
All other Sinclair products in stock. 2,000 amplifier, \$236. Carr. 7/6. Neoteric amplifier \$48, Carr. 7/6.

* TRANSISTORISED FM TUNER *



6 TRANSISTOR
HIGH QUALITY
TUNER, SIZE
ONLY 6 4×21in.
3 J.F. stages.
Double tuned dis-

on 9V battery, Coverage 88-108 MHz, Ready built ready for use, Pantastic value for money, 28.7.8, P. & P. 2/8, Stereo multiplex

Our latest edition giving full details of a comprehensive range of HI-FIEQUIPMENT COMPONENTS, TEST EQUIPMENT and COMMUNICATIONS EQUIPMENT, FREE DISCOUNT COUPONS VALUE 10/-, 248 pages, fully illustrated and u thousands of items at bargain prices detailing



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P. & P. 1/6

G. W. SMITH & CO. (RADIO) LTD. See previous page See opposite page \rightarrow

RUSSIAN C1-16 DOUBLE BEAM OSCILLOSCOPES

5MHz Pass Band. Separate Y1, Y2 amplifiers. Calibrated triggered sweep from

0-2µsec to 100msec/cm.
Supplied complete with all accessories and instructions, \$87, Carr. paid.



MARCONI CT44/ TF956 AF WATTMETER

1 μ/watt to 6 watts. 220. Carr. 20/-.

DECADE RESISTANCE ATTENUATOR ब्बंबंबं 🚳 🚳 Variable range 0-111dB. Con-

0-111dB. Connections. Unbalanced T and Bidge T. Impedance
600Ω range (0-1dB::10) + (1dB::10)
+ 10 + 20 + 30 + 40dB. Frequency:
d.c. to 200kHz (-3dB). Accuracy:
0-03dB. +Indication dB × 0-01. Maximum input less than 4W (50V). Built in
600Ω load resistance with internal/external
switch. Brand new \$27.10.0. P. & P. 5/-

BELCO AF-5A SOLID STATE SINE SQUARE WAVE C.R. OSCILLATOR



Sine 18-200,000Hz: Square 18-50,000Hz. Output max +10dB ($10k\Omega$). Operation internal batteries.

Attractive two-tone case 7gin : 5 Price £17,10.0 Carr. 3/6. Sin 2in.



TE-16A Transistorised
Signal Generator, 5 ranges
400 kHz-30 mHz. An
inexpensive instrument
for the handyman. Operates on 99 battery. Wide,
easy to read scale,
800 kHz modulation.
5½ > 5½ > 31½. Complete
with instructions and
leads. \$7.19.6. P.&P. 4/-.

BELCO DA-20 SOLID STATE DECADE AUDIO OSCILLATOR



New high quality portwhich is the strument, Sine1Hz to 100kHz. Square 20Hz to 20kHz. Output max. +10dB (10kΩ). Operation 220/240V a.c. Size 0V a.c. 150mm tion 220/240V a.c. Size 215mm : 150mm : 120mm : Price **£27,10.0**, Carr.

T.E.40 HIGH SENSITIVITY

A.C. VOLTMETER
10 meg. input 10 ranges:
-01 /-003 /-1 /-3 /-1 /
3 / 10 / 30 / 100 / 300 V.
R.M.S. 4c/s.-12Mc/s.
Decibels -40 to +50dB.
Supplied brand new
complete with leads and
instructions. Operation
230V a.e. \$17,10.0. Carr. 5/-.



TE-65 VALVE VOLTMETER



High quality instrument with 28 ranges. D.c. volts 1.5-1,500V. A.c. volts 1.5-1,500V. Resistance up to 1,000 megohms, 220/240V a.c. operation. Complete with probe and instructions. 217.10.0 P. & P. 6/-. Additiona Probes available: R.F 42/6. H.V. 50/-. £17 10 0

AUTO TRANSFORMERS 0/115/230v. Step up or step down. Fully

led. 150 W. 22.7.6, P. & P. 3/6 300 W. 23.5.0, P. & P. 4/6 500 W. 24.19.6, P. & P. 6/6 1,000 W. 27.5.0, P. & P. 7/6 1,500 W. 23.19.6, P. & P. 8/6 7,500 W. 215.10.0, P. & P. 20/-

SOLID STATE VARIABLE A.C. VOLTAGE REGULATORS



E REGULATORS
Compact and panel mounting. Ideal for control of lamps, derills, electrical appliances, etc. Input 250/240V a.c. Output continuously variable from 29V to 230V.
Model MR2305 5A 68 × 46 × 48mm, 28.7.8, Model MR2310 10A 30 × 68 × Pontage 2/6.

60mm, £11,19,6 Postage 2/6.

MULTIMETERS for EVERY purpose



TECH PT-34, 1,000 O.P.V. 0/10/50/250/ 500/1,000V a.c. and d.c. 0/100K, 39/6, P. & P. 2/6.



MODEL TE-200 | MODEL TE-200, 20,000 |
| O.P.V. Mirror scale, 90,000 |
| O.P.V. Mirror scale, 90,000 |
| 12511,000 v d.c. 0/15/25/1 |
| 250M.λ. 0/60K/β | neg. |
| 290M.λ. 0/60K/β | neg. |
| 120 to •62dB. 75/- v. & P. 3/-



MODEL 500. 30,000 O.F.V. with overload protection, mirror scale 0/-5/2-5/10/25/100/ 250/5001/000 V d.c. 0/25/ 10/25/100/25/505/ 10/25/100/25/505/ 10/25/100/25/505/ 500 nt. 12 amp. d.c. 0/60K/6 Meg./60 Meg.Ω. 28,17.8. Post paid.



MODEL TE-70, 30,000 O.P.V. 0/3/15/60/300/ 600/1,200V d.c. 0/6/30/ 120/600/1,200V a.c. 0/ 30µA/3/30/00mA, 0/ 16K/160K/1-6M/16 meg.

TMK MODEL TW-50K. 46 ranges, mirror scale. 50K/Volt d.c. 5K/Volt d.c. 5K/Volt d.c. D.c. volts: 0-125, 0-25, 1-25, 2-5, 5, 10, 2-3, 50, 126, 266, 300, 1,000V. D.c. current: 25, 50LA, 2-5, 5, 2-5, 5, 5, 5, 5, 50, 250, 500mA, 5, 10A. Resistance: 10K, 100K, 1 meg., 10 meg. Decibles: -20 to +81-50B, \$8.17.6, P. & P. 3/6.



TE-900 20,000 Ω/VOLT GIANT MULTIMETER.

GIANT MULTIMETER.

Wirror scale and overload protection. 6in full view meter. 2 colour scale. 0/

V n.c. 0/25/12-5/10/50/250/1,000/
V d.c. 0/30/4/110/100/300mA/J104
02/K/200/K/20 meg. ohm. \$15,0.0. P. & P. 5/



Maria.

504A

MODEL 5025, 57 ranges, giant 51in meter, polarity reverse switch. Sen-

reverse switch. Sensitivity: 36K/Yott a.c. ak/Yott a.c. a 100K, 1 meg, 10 meg. Decibels: +85dB, 212.10.0, P, & P, 3/6.

AVOMETER MOVEMENTS

HONOR TE 10A. HONOR TE.10A. 20k \(\Omega \)/
Volt \(2\)/26/50/26/50/60/
2.500V d.c. \(10\)/50/100/500/
1.000V a.c. \(0\)/50/26/56 mA
250lnA d.c. \(0\)/6K/\(\Omega \) mer.
0hm. \(-20 \) to \(+22\)dB.
10-0, 100 \(\Omega \) mtd. \(0\)-100-0-1 mtd. \(\Omega \)/6K/\(\Omega \) P. \(\omega \) P 3/2



MODEL TE-300, 30,000 O.P.V. Mirror scale, over-load protection 0/0-6/3/16/ 60/300/1,200V a.c. 0/6/30/ 120/600/1,200V a.c. 0/ 30µA / 6nnA / 60nnA / 300nnA/600nnA . 0/8K/80K/800K/8 n -20 to +63dB. 25.19.8, P. & P. 3/-

MODEL TE 80. 20,000

MODEL TE 80. 29,000 0.P.V. 0 / 10 / 50 / 100 / 500 / 500 / 1.000V. a.c. 0/5/25/50/ 250 / 500 / 1,000V d.c. 0.5µA. 5/50/500mA./ 0/6K/60/K/600K/6 Meg. 24.17,6. P. & P. 3/-



27.10. F. & F. 3]-.

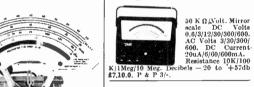
TMK MODEL TW-90CB.
Features Resettable Overbond Button. Sensitivity:
20 K Ω/Volt d.c. 5 K Ω/
Volt n.c. D.c. volts: 0-0-5,
2-5, 10, 50, 250, 1,000V.
A.c. volts: 0-2-5, 10, 60, 250, 1,000V.
D.c. currents: 0-0-05, 5, 5, 50, 500mA, 10A.
Resistance: 0-5 K, 50K, 0-500 K, 6 meg.
Decibels: -20 to +52dB. #11,10.0. P. &
P. 3/6.

MODEL AS-10QD, 100K Ω / Volt. 5in, nirror scale.
Bullt-in meter protection 0/
3/12/60/120/300/600/1,200V
d.c. 0/6/30/120/300/600V
a.c. 0/10µA/6/60/300nA/
12A. 0/2K/200K/2M/200M.
+17dB. \$12,10,0_R. & P. 3/6.





SKYWOOD SW-500



Spare movements for Model 8 or 9. (Fitted with Model 9 scale) or basis for any multimeter, Brand New Boxed 69/8, P. & P. 3/6.

-2010

270° WIDE ANGLE

1mA METERS

MW1-6 60mm square 23,19,6, MW1-8 80mm square 24,19,6, P. & P. extra.



"YAMABISHI" VARIABLE VOLTAGE TRANSFORMERS

MODEL S-260

General Purpose Bench Mounting

10.

Excellent quality . Low price . Immediate delivery MODEL S-260 B **Panel Mounting**

> 2.5A Please add Please add postage
> ALL MODELS
> INPUT 230 VOLTS, 50,60
> CYCLES, OUTPUT
> VARIABLE 0-260 VOLTS
> Special discounts for quantity

UNR-30 RECEIVER



4 Bands covering 550KHz-30MHz. B.F.O. Built in Speaker 220/240V a.c. Brand new with instructions. 215.15.0. Carr. 7/6.

WS62 TRANSCEIVERS Large quantity available for EXPORT! Excellent condition. Enquiries invited.

UR-IA SOLID STATE



4 Bands covering 50KHz 30MHZ, FET, 8 Meter, Variable BFO for 88B. Built in Speaker, Telescopic Aerial, Bandspread, Sensitivity Control. 220/240V a.c. or 12V d.c. 122in 4in 7in. Brand new with instructions. 225. Carr. 7/6.

LAFAYETTE HA-600 RECEIVER



General coverage 150-400KHz, 550KHz, 30MHz. FET front end, 2 mech. filters, product detector, variable B.F.O., noise limiter, 8 Meter, Bandspread. RF Gain. 15in · 94in × 84in. 18ib. 220/240V a.c. or 12V d.c. Brand new with instructions. \$45. Carr. 10/-.

LAFAYETTE "
HA-800 SOLID
STATE AMATEUR COMMUNICATION RECEIVER



3-5-4, 7-7-3, 14-14-35, 21-21-45, 28-29-7, 50-54MHz. Dual conversion, 2 mech filters, product detector, variable BFO. 8 Meter, 100KHz calibrator. 220/246V a.c. or 12V d.c. 15in. 9 fin. 18lin. 18lin.



LAFAYETTE PP60 VHF FM RECEIVER

Solid State. 132-174MHz. Fully tuneable or crystal controlled (not supplied). Built in Speaker, Squelch and Volume Controls. 220/240V a.c. or 12V d.c. Brand new with instructions. \$37,10.0. Carr. 10/-.

FULL RANGE OF TRIO EQUIPMENT

VOLTAGE STABILISER TRANSFORMERS 180-260V input. Output 230V. Av 150W or 225W. 212,10.0. Carr. 5/-.

Available

EDDYSTONE VHF RECEIVERS MODEL 770R. 19-165 Mc/s. Excellent condition. #150.

INTERCOM/BABY SITTER



Transistorised conis, ideal for home office / work-shop, etc. 2-way buzzer call system. For desk or wall mounting. Supplied complete with connecting wire, bat-Transistorised coms, ideal

teries, instructions, 2 station 59/6. P. & P. 2/6. 4 station 26,12,6. P. & P. 5/-.

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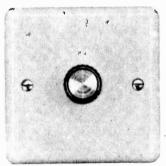
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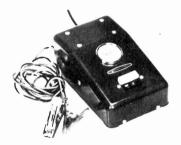
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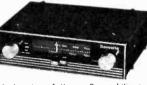
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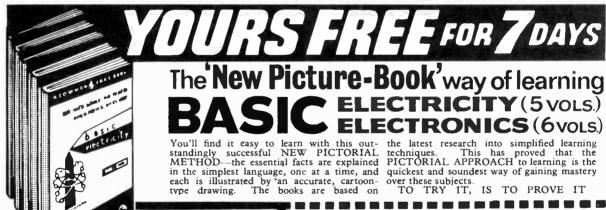
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resistance 0-60kg; decibels -20 to +22dB. Size of meter 4in×3in×1in. Complete with leather case. 85/-, P. & P. 3/6.

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Listen to all your favourite radio programmes anywhere and anytime. Lightweight and comfortable to wear. Covers full medium waveband, supersensitive tuning control and variable volume control. Ideal for out and indoors. Only \$10.19.6. P. & P. 4/6.

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Same size as standard wall switch and will dim incandescent lighting from full on to off. Heavy plastic box with knob. Lind-Air Price 24.10.0. P. & P. 3/6.



Comfortable on, with trigger control, "U" shaped hit to 3\frac{1}{3}\text{in.} bit to minimise wear. Light beam is automatically directed on to end of bit when ON/OFF trigger is in use. 230-250 volts. 85 watt plement. 49/6, P. & P. 3/6.

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Panels, Jigs and Accessories and the book-let "Projects on S-DeC" all contained in a strong attractive plastic case. Ideal for the professional user. 25.17.6. P. & P.

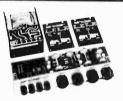
T-DeC KIT 22.10.0. P. & P. 3/6.

62 D. MULTI-TESTER

29,000 ohms per volt; de. voltage 6, 25, 50, 250, 600 V, 2-5kV (29,000 ohms per volt), a.e. voltage 10, 50, 100, 500, 1,000 v (10,000 ohms per volt); d.e. current 0-50μA, 0-2-50mA; nesistance 0-6kΩ, 0-6MΩ (300 ohms and 30kΩ at centre scale); capacitance 10F to 0-001μF to 0-01μF (collebles -20 to 4-22d B. Size 4 in × 3 in × 1 in. 71/-, P. & P. 3/6. 20,000 ohnis ner volu



60



An attractive alternative for the cuthusiast prepared to assemble these excellent modules to make a stereo assembly. Z.30 24W Power Amplifier 89/6 (2 required). Stereo Sixty Control/Pre-amplifier 89,19.6. P.Z.5 Power Supply Unit \$4.19.6. SINCLAIR PROJECT 60 Package deal price 19 Ong. P. & P. 12/6. Project 60 is supplied complete with instruction manual and templates for plinth mounting.

SINCLAIR IC-10 INTEGRATED CIRCUIT 10W Amplifier. Size only lin × 0-4in × 0-2in. A true hi-fi amplifier complete with manual giving details of a wide range of applications and instructions. Guaranteed 0.2in.

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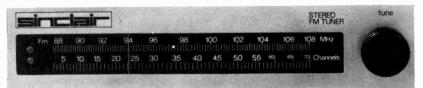
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Project 60







the world's most advanced high fidelity modules

With the introduction of an entirely new and original high fidelity stereo F.M. tuner, the Project 60 range can be said at this stage to be complete. It offers the constructor a most attractive choice of modular arrangements whereby a high fidelity system can be selected to suit the user's personal requirements. Equally, it is possible to use any Project 60 modules separately or partially grouped and so benefit greatly from the flexibility in use these modules afford. The chart below shows somr of the most popular applications for constructors to assemble. The Project 60 manual (free with the modules) suggests others as well and its 48 pages are packed with valuable information. The new tuner, for example can be used with any good high fidelity system as well as Project 60.

Project 60 now falls into four interdependent groups: – 1. The Z.30 and Z.50 amplifiers which have only 0.02% distortion at all output levels and are useful in a wide variety of other applications. 2. The control units comprising the Stereo 60 preamp and control unit and the Active Filter Unit (A.F.U.) with which both high pass and low pass filtering can be introduced between control unit and power amplifiers. 3. The Stereo F.M. tuner as described opposite; and 4. The power supply units PZ.5.

PZ.6 and PZ.8. For most requirements when using Z.30 power amplifiers, the PZ.5 will be perfectly adequate; if low efficiency (high quality) loud speakers are used, the PZ.6 stabilised power supply unit will be used. The PZ.8 will be needed with Z.50s which can be used for any Project 60 system.

Project 60 modules incorporate some of the most advanced circuitry in the world to achieve unsurpassed standards of high fidelity and modern manufacturing techniques enable these modules to be sold at exceptionally attractive prices. Assembling the modules requires no skill or previous experience since the manual supplied with the modules explains clearly how everything can be done with nothing more than the simplest of domestic tools.

Project 60 manuals

How to assemble and use Project 60 modules to best advantage in the above and other applications will be found in the fully descriptive Project 60 manual included with Project 60 systems. This 48 page manual is available separately, price 2/6d including postage.

	System	The Units to use	In conjunction with	Cost of Units	+ Project 60 tuner
A	Car Radio .	Z.30	Existing carradio, Sinclair Micromatic	89/6	
В	Simple battery powered record player	Z.30	Crystal pick-up, 12V or more battery supply and volume control	89/6	· · · · · · · · · · · · · · · · · · ·
C	Mains powered record player	Z.30 and PZ.5	Crystal or ceramic P.U. Volume control etc.	£9.9.0	£34.9.0
D	20+20 watts R.M.S. stereo amplifier for most needs	Two Z.30s, Stereo 60 and PZ.5	Crystal, ceramic or magnetic P.U., most dynamic speakers, F.M. tuner etc.	£23.18.0	£48.18.0
E	20+20 watts R.M.S. stereo amplifier for use with low efficiency (high performance) speakers	Two Z.30s, Stereo 60 and PZ.6	High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc All dynamic speakers	£26.18.0	£51.18.9
F	40+40 watts R.M.S. de-luxe stereo amplifier	Two Z.50s, Stereo 60 PZ.8 and mains transformer	As for E	£32.17.6	£57.17.6
G	Outdoor public address system	Z.50	Microphone, up to 4 P.A. speakers. 12V car battery with converter, or 45V d.c., controls	£5.9,6	
H	Indoor P.A.	One Z.50, PZ.8 and mains transformer	Microphone, guitar, heavy duty speakers etc., controls	£17.8.6.	
J	High pass and low pass filters	A.F.U.	D, E or F as above	£5.19.6	



Sinclair Radionics Limited, 22 Newmarket Road, Cambridge
Telephone (0223) 52731

Z.30 & Z.50 power amplifiers

The Z.30 together with the Z.50 are both of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low 0.02% at full output and all lower outputs. Whether you use the Z.30 or Z.50 power amplifiers in your Project 60 system will depend on personal preference, but they are the same physical size and may be used with other units in the Project 60 range equally well. For operating from mains, for the Z.30 use PZ.5 for most domestic requirements, or PZ.6 if you have very low efficiency loudspeakers. For Z.50, use the PZ 8 described below.

SPECIFICATIONS (Z.50 units are interchangeable with Z.30s in all applications). Power Outputs

Z.30 15 watts R.M.S. into 8 ohms, using 35V: 20 watts R.M.S. into 3 ohms using 30 volts.
Z.50 40 watts R.M.S. into 3 ohms from 40 volts:

30 watts R.M.S. into 8 ohms, using 50 volts. Frequency response 30 to 300,000 Hz ± 1dB Distortion 0.02% into 8 ohms

Signal to noise ratio better than 70 dB unweighted Input sensitivity 250mV into 100 Kohms. For speakers from 3 to 15 ohms impedance. Size $3\frac{1}{2} \times 2\frac{1}{4} \times \frac{1}{2}$ ins.

Stereo 60 pre amp/control unit

Designed for the Project 60 range but suitable for use with any high quality power amplifier. Again silicon epitaxial planar transistors are used throughout. achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

SPECIFICATIONS

- •Input sensitivities Radio up to 3mV, Mag. p.u. 3mV: correct to R.I.A.A. curve ± 1dB: 20 to 25,000Hz, Ceramic p.u. – up to 3mV : Aux. – up to 3mV.

 Output – 250mV.
- Signal-to-noise ratio better than 70dB
- Channel matching within 1dB.
 Tone controls TREBLE +15 to —15dB at 10kHz: BASS +15 to —15dB at 100Hz.

Built, tested and guaranteed with circuits and

Built, tested and guaranteed with circuits and instructions manual

Front panel – brushed aluminium with black knobs

Size 8½ x 1½ x 4 ins

7.30

instructions manual

Built, tested and guaranteed

£9.19.6

Z.50

Z.30

89/6

109/6

Active Filter Unit

For use between Stereo 60 unit and two Z.30s or Z.50s, the Active Filter Unit matches the Stereo 60 in styling and is as easily mounted. It is unique in that the cut-off frequencies are continuously variable, and as attenuation in the rejected band is rapid (12dB/octave), there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The Sinclair A.F.U. is suitable also for use with any other amplifier system.

Two stages of filtering are incorporated - rumble (high pass) and scratch (low pass). Supply voltage – 15 to 35V. Current – 3mA. H.F cut-off (–3dB)

variable from 28kHz to 5kHz, L.F cut-off (-3dB) variable from 25Hz to 100Hz, Filter slope, both sections 12dB per octave. Distortion at 1kHz (35V supply) 0.02% at rated output.

Built, tested and quaranteed

£5.19.6

Power Supply Units

The units below are designed specially for use with the Project 60 system of your choice.

Illustration shows PZ.5 power supply unit to left and PZ.8 (for use with Z.50s) to the right. Use PZ.5 for normal Z.30 assemblies and PZ.6 where a stabilised supply is essential.

PZ-5 30 volts unstabilised £4.19,6 PZ-635 volts stabilised £7.19.6 PZ-8 45 volts stabilised

(less mains transformers) £5.19.6 PZ-8 mains transformer £5.19.6

GUARANTEE If within 3 months of purchasing Project 60 modules directly from us, you are dissatisfied with them, we will refund your money at once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service it at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail, Air-mail charged at cost.



Stereo FM tuner



first in the world to use the phase lock loop principle

Before production of this tuner, the phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio over other systems. Now, for the first time the principle has been applied to an FM tuner with fantastically good results. By the inclusion of other original features such as varicap diode tuning, printed circuit coils and an I.C. in the specially designed stereo decoder, the tuner has an unsurpassed specification, which also incorporates a squelch circuit for silent tuning between stations, A.F.C. and A.G.C. Sensitivity is such that good reception becomes possible in difficult areas, foreign stations can be tuned in suitable conditions and often a few inches of wire are enough for an aerial. In terms of high fidelity, this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. Although the tuner is intended primarily for use with a Project 60 system, it can be used to advantage with any other high fidelity system. It is easily mounted into any cabinet as shown in the manual supplied with it.

Specifications

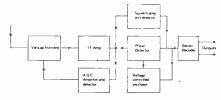
Number of transistors 16 plus 20 in I.C. Tuning range 87.5 to 108 MHz Capture ratio 1.5dB Sensitivity 2µV for 30dB quieting 7μV for full limiting

Squelch level 20µV A.F.C. range ± 200 KHz Signal to noise ratio > 65dB

Audio frequency response 10Hz-15kHz(±1dB) Total harmonic distortion 0.15% for 30% modulation

Stereo decoder operating level 2µV Pilot tone suppression 30dB Cross talk 40dB I.F. frequency 10.7 MHz Output voltage 2 x 150mV R.M.S. Aerial Impedance 75 Ohms Indicators Mains on ; Stereo on ; tuning indicator

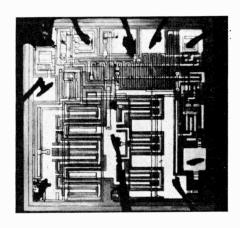
Operating voltage 25-30 VDC Size 3.6 x 1.6 x 8.15 inches: 91.5 x 40 x 207 mm

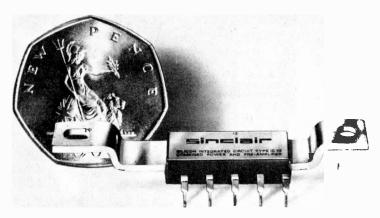


Price: £25 built and tested. Post free.

To: Sinclair Radionics Ltd., 22	Newmarket Road, Cambridge
Please send	NAME
	ADDRESS
for which I enclose cash/cheque money	
order	

Sinclair IC-10





the world's most advanced high fidelity amplifier

Specifications

Output: 10 Watts peak, 5 Watts R.M.S. continuous Frequency response: 5 Hz to 100 KHz±1dB Total harmonic distortion: Less than 1% at full output. 3 to 15 ohms. Load impedance: Power gain: 110dB (100,000,000,000 times) total. 8 to 18 volts. Supply voltage: 1 x 0.4 x 0.2 inches. 5mV. Sensitivity: Input impedance: Adjustable externally up to 2.5 M ohms.

Circuit Description

The first three transistors are used in the pre-amp and the remaining 10 in the power amplifier. Class AB output is used with closely controlled quiescent current which is independent of temperature. Generous negative feedback is used round both sections and the amplifier is completely free from crossover distortion at all supply voltages, making battery operation eminently satisfactory.

Applications

Each IC-10 is sold with a very comprehensive manual giving circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include stabilised power supplies, oscillators, etc. The pre-amp section can be used as an R.F. or I.F. amplifier without any additional transistors.

The Sinclair IC-10 is the world's first monolithic integrated circuit high fidelity power amplifier and pre-amplifier. The circuit itself, a chip of silicon only a twentieth of an inch square by one hundredth of an inch thick, has 5 watts R.M.S. output (10w. peak). It contains 13 transistors (including two power types), 2 diodes, 1 zener diode and 18 resistors, formed simultaneously in the silicon by a series of diffusions. The chip is encapsulated in a solid plastic package which holds the metal heat sink and connecting pins. This exciting device is not only more rugged and reliable than any previous amplifier, it also has considerable performance advantages. The most important are complete freedom from thermal runaway due to the close thermal coupling between the output transistors and the bias diodes and very low level of distortion.

The IC-10 is primarily intended as a full performance high fidelity power and pre-amplifier, for which application it only requires the addition of such components as tone and volume controls and a battery or mains power supply. However, it is so designed that it may be used simply in many other applications including car radios, electronic organs, servo amplifiers (it is d.c. coupled throughout), etc. Once proven, the circuits can be produced with complete uniformity which enables us to give a full guarantee on every IC-10, knowing that every unit will work as perfectly as the original and do so for a lifetime.

IC-10

with IC-10 manual Post free. 59/6

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	ADDRESS
for which I enclose cash cheque money	PE/171.8



Q.16 High fidelity loudspeaker

Developed out of the revolutionary and much praised design of the original Sinclair Q.14 comes this more advanced version to meet the requirements of even greater numbers of high fidelity enthusiasts. The Q.16 employs the same well proven acoustic principles in which a special driver assembly is meticulously matched to the physical characteristics of the uniquely designed housing. In reviewing this exclusive Sinclair design, technical journals have been loud in their praise for it and it comfortably stands comparison with very much more expensive loudspeakers. The shape of the Q.16 enables it to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures, and with its improved styling, the Q.16 presents an entirely new and attractive appearance. A solid teak surround is used with a special all-over cellular black foam front chosen as much for its appearance as for its ability to pass all audio frequencies unimpaired.

The Q.16 is compact and slim and is the ideal shelf-mounted speaker, and brings genuine high fidelity within reach of every music lover

Specifications

Loading:

Construction: A sealed seamless sound or pressure chamber is used with internal baffle, all of

materials carefully chosen to ensure freedom from spurious tone coloration.

Up to 14 watts R.M.S

Input impedance: 8 ohms

Frequency response: From 60 to 16,000Hz, as confirmed.

by independently plotted B & K curve.

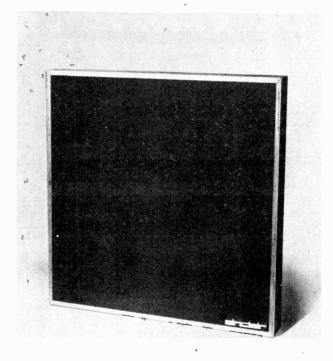
Driver unit: Specially designed high compliance unit

having massive ceramic magnet of 11,000 gauss, aluminium speech coil and special cone suspension. Excellent transient response is achieved

Size and styling: $9\frac{3}{4}$ " square on face \times $4\frac{3}{4}$ " deep with neat

pedestal base. Black all-over cellular foam front with natural solid teak surround.

Price: £8 19 6.



Micromatic Britain's smallest radio

Considerably smaller than an ordinary box of matches, this is a multi-stage A.M. receiver meticulously designed to provide remarkable standards of selectivity, power and quality. Powerful A.G.C. is incorporated to counteract fading from distant stations; bandspread at higher frequencies makes reception of Radio 1 easy at all times. Vernier type tuning plus the directional properties of the self-contained special ferrite rod aerial makes station separation very much easier than with many larger sets. The plug-in high fidelity type magnetic earpiece which matches exactly with the output of the Micromatic provides wonderful standards of reproduction both for speech and for music. Everything including the batteries is contained within the attractively designed case. Whether you build your Micromatic or buy it ready built and tested, you will find it as easy to take with you as your wristwatch, and dependable under the severest listening conditions

Specifications

Earpiece:

Battery requirements:

Controls

Case

Size: $1\frac{1}{18}'' \times 1\frac{7}{78}'' \times \frac{1}{2}''$ (46 × 33 × 13mm). Weight including 1 oz. (28.35gm) approx.

batteries:
Tuning: Medium wave band with bandspread at

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High-fidelity magnetic type.
Two Mallory Mercury Cells, type R.M. 675,

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Black plastic with anodised aluminium

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No. 1

ELECTRONICS

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SEEING WITH SOUND

ALMOST two whole years have passed since fog was last discussed on this page. More specifically, we then aired our hopes for an electronic system that would, to some degree, restore visibility to drivers of motor vehicles in bad weather conditions.

Two years is a long time in electronics, and one could quite reasonably expect important steps to have been taken by now toward the solution of this serious problem. But nothing concrete has emerged so far from industry or from government research establishments.

We do know that much effort has been put into the development of a radar system for vehicles operating inside airports. There is a vital need for emergency service vehicles to be able to move at speed to the scene of an aircraft mishap, no matter what state of visibility obtains. This particular radar system has not yet been put into production, presumably because the airport authorities have not yet been decided to buy.

By normal standards, a radar system is necessarily expensive and involves much sophisticated engineering. It is true that solid state devices such as Gunn-effect diodes are available for microwave operation, and these help to simplify the receiver and transmitter design and so reduce the overall bulk of the installation. But there still remains a major problem in the form of the display device. Until the c.r.t. is superceded by a solid state electro-luminescent panel, there is little likelihood that radar, even in an elementary form, can be seriously considered as a feasible proposition for installation on all types of road vehicles.

While waiting for further advances in technology to solve these engineering (and cost) problems, there is no reason why private individuals should not on their own account investigate other methods to combat the common enemy, fog.

Leaving aside electro-magnetic waves, why not explore the possibility of sound waves? The equipment needed for an acoustical system is not highly complicated, and the basic principles involved are familiar enough to most electronics enthusiasts. The Sonic Obstacle Locator described in this issue has proved itself to a marked degree. Admittedly, as a driving aid it is not infallible, and lacks, as yet, the ability to discriminate between tree, lamp post, or car. Serious limitations, in fact, for a car-borne device. But it is not offered as an instant solution to a formidable problem. It is offered as an idea worthy of further development by the experimentally inclined. And other uses may well become apparent.

We hope many will rise to the challenge. Amateurs have blazed the trail before. Most significantly, in proving the usefulness of the then neglected high frequency radio waves. Maybe private experimenters can achieve some noteworthy success in electro-acoustics.

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ATMOSPHERIC pollution and weather conditions often combine to produce thick fogs which reduce visibility to as little as 10–20 feet. Much the same applies where there is smoke or driving snow. In such conditions man's senses are virtually useless and there is a real need for technical aids to penetrate the murk.

Short range radar would seem to provide an answer to the problem, but the equipment is expensive and involves radio licensing complications. Fog is transparent to certain bands of infra-red, but infra-red emitters and sensors for use at such wavelengths are not, as yet, sufficiently developed. Another approach would be ultrasonics, but high frequency sound waves tend to be scattered by fog droplets, and have a short range.

AUDIBLE SOUND

The Oil-bird of South America points the way to a possible solution to the problem; it finds its way around dark caves by emitting audible clicking noises, like the sound of castanets. Audible sound is, of course, only slightly attenuated by fog and other atmospheric particles, hence the use of the fog horn, and the explosive cap on railways.

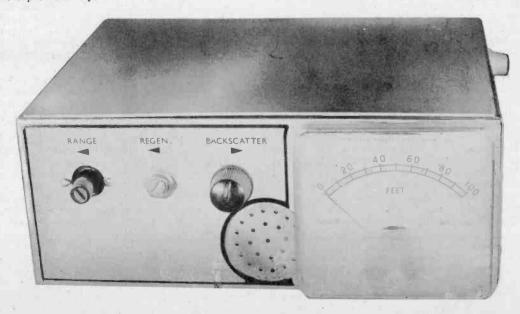
Sonar has long been employed for navigation and ranging purposes underwater, and there seems to be no reason why audible sound pulses in air should not be similarly used if the equipment can be made to operate against a background of day-to-day ambient noise.

The Sonic Obstacle Locator is the result of experiments to test the practicability of using audible sound to detect the presence of unseen objects in fog. With a range of 100 feet, the equipment could be used on board small boats for docking in conditions of poor visibility, or as an outdoor, all weather intruder alarm. There are many other possible applications, such as counting passing cars on a roadway, and educational demonstrations involving the speed, refraction, and reflection of sound waves.

SENSITIVITY

Although not recommended for use on cars in fog, the Sonic Obstacle Locator will work at low speeds, and was tested on a vehicle as this provided the most adverse conditions of high ambient noise.

To give an idea of sensitivity, a response is obtained from an adult pedestrian at up to 22 feet, a small parked car at 50 feet, or a stone wall perpendicular to



LOCATOR By D. Bollen

the sound source axis at 100 feet. When an obstacle in the road ahead is detected, the unit inside the car emits a bleep warning note, of duration directly proportional to distance.

The note varies from a short click at 95 feet range, to an almost continuous tone at the minimum range of 5 feet. A meter linearly calibrated 0-100 feet is incorporated to show the actual distance of the obstacle.

Because of sonic interference from the exhaust noise of other vehicles, particularly diesel lorries, the equipment will only operate reliably at traffic speeds below 30 m.p.h., under typical fog driving conditions.

MEASURING DISTANCE THROUGH AIR WITH SOUND

High frequency sound is readily reflected from solid surfaces, and sound waves conform to the same general laws of reflection as light.

The time taken for a sound to travel to a reflecting surface and back again can be used as a measure of the intervening distance, with an accuracy dependent on slight changes in the speed of sound resulting mainly from variations of temperature and humidity.

Sound velocity is not significantly influenced by droplets or particles in the atmosphere, and is largely independent of air pressure and altitude when corrections are made for temperature.

The velocity of sound in dry air at 0 degrees centigrade is 1,087 feet per second; about 4½ times slower than sound in water. If the air temperature is raised to 20 degrees centigrade a velocity increase of about 3.7 per cent will be noted.

When the air is fully saturated with water vapour, this causes a velocity increase of about 0.5 per cent, because water vapour decreases the density of air.

From the above it can be estimated that, over a temperature range ±10 degrees centigrade, and in varying humidity, a sonic yardstick should offer a potential accuracy of about ± 2.5 per cent.

OPERATING FREQUENCY AND **TRANSDUCERS**

The frequency of the transmitted sound should be high, to combat interference from man-made noise, the bulk of which lies in the lower part of the audio spectrum. However, sonic transducers operating above 15kHz are usually not sensitive enough for low power sound-through-air applications, although they may be suitable for service underwater where sound conduction

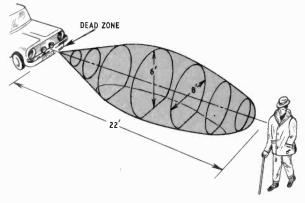


Fig. 1. Area of response of the Sonic Obstacle Locator when used for pedestrian detection

is good. In the interests of economy, it was decided that special transducers should be avoided.

Tests were made with several types of loudspeaker and microphone, to establish a suitable compromise between cost and noise rejection when choosing the operating frequency. Firstly, the upper limit was set by microphones offering a useful output above 15kHz, as they were too expensive to qualify.

It was discovered that good quality tweeters operating at 10-15kHz gave markedly inferior results when compared with small, cheap loudspeakers handling 10kHz. This somewhat surprising result was eventually attributed to sound reinforcement arising from resonances in the cheap loudspeakers, and further improvements at 10kHz were effected by cutting down the size of the loudspeaker cone, and stiffening it with cellulose dope. Obviously, if the sonic transmitter can be encouraged to provide a large acoustic output, this will tend to minimise noise, and compensate for the lower frequency.

When a loudspeaker is handling frequencies above a few kilohertz there will be only a small excursion of the voice coil and cone, and the limit to the amount of power the loudspeaker can handle is determined by voice coil overheating. If the high frequency input is pulsed on and off, with a relatively long pause between pulses, the loudspeaker can sustain very high levels of input without damage, and at the same time yield an

output of high intensity.

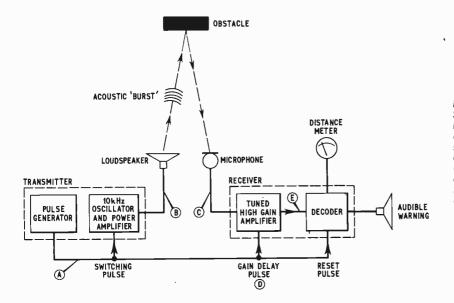


Fig. 2. Block diagram of the Sonic Obstacle Locator. 10kHz bursts of radiated acoustic energy are made to bounce off the obstacle. These are then amplified and the distance of the obstacle computed by the decoder for display by a meter. An audio bleep warning is also triggered at this final stage

DIRECTIONAL CHARACTERISTIC

In using a locator, echoes should only be obtained from obstacles in the road ahead, and not from objects which do not lie in the path of the vehicle.

A loudspeaker offers a polar response, at high frequencies, which is directional, but is not sufficiently narrow in shape for the present application; the requirement taking the form of a broad sphere slightly flattened along the sound source axis. However, a microphone can be made to see only a narrow angle by placing it at the focus of a parabolic reflector which looks along the source axis.

When loudspeaker and microphone polar responses are combined, the result is a thin, almost ideal cigar shape extending in front of the vehicle, as depicted in Fig. 1.

The plot upon which Fig. 1 is based was obtained from a 5 gallon oil drum which simulates a pedestrian wearing normal, soft clothing.

It can be seen that the area of response is roughly the width of the car and extends forwards to a distance of 22 feet. For larger obstacles, the area of detection will be wider and longer, though still of the same cigar shape.

SONIC TRANSMITTER

The basic features of the Sonic Obstacle Locator are outlined by the block diagram of Fig. 2, with associated waveforms in Fig. 3.

Dealing first with the transmitter, this consists of a pulse generator, a 10kHz square wave oscillator, and a tuned power amplifier. The pulse generator provides a master pulse of 2ms duration (Fig. 3A) at intervals of 178ms, the time taken for sound to travel to the obstacle and back at the maximum range of 100 feet.

On receipt of the pulse, the 10kHz oscillator provides an output of 2ms duration, consisting of 20 cycles of a 10kHz square wave, which is then passed to the power amplifier. Being tuned, the power amplifier converts the square wave into a sine wave, and the input to the loudspeaker is an envelope containing 20 cycles of a 10kHz sine wave as seen in Fig. 3B.

To detect echoes from obstacles very close to the transmitter, the radiated sound must be of short duration, so that it dies away completely before the

echo arrives. A time of 2ms for the acoustic burst from the transmitter is well inside the period fixed by the minimum range of 5 feet, to allow for various circuit delays and loudspeaker lag.

In passing, it is of interest to note that the short burst of 10kHz sound from the transmitter is registered by the human ear merely as a loud click, with virtually no tone content.

RECEIVER

If an untuned amplifier was used with a microphone to pick up the minute reflected signals from the transmitter, the echo pulses would be completely swamped by

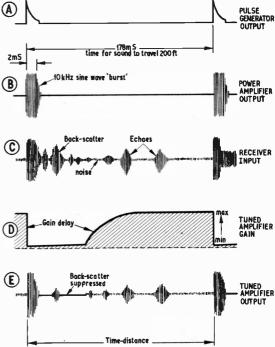


Fig. 3. Waveforms associated with Fig. 2 and identifiable by the letter. Fig. 3D is the tuned amplifier gain characteristic

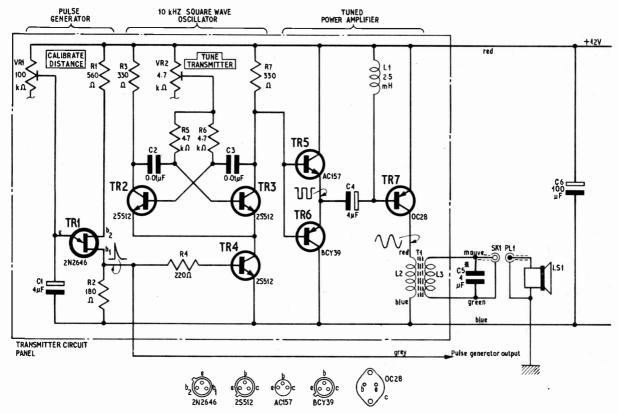


Fig. 4. Circuit diagram of the transmitter consisting of pulse generator, square wave oscillator and tuned power amplifier

broadband ambient noise, except in a very quiet environment. Therefore, the sonic receiver is tuned to the transmitter frequency. In the block diagram the receiver is seen to consist of a microphone, a tuned high gain amplifier, and an echo pulse decoder circuit.

A typical microphone response to the burst of sound from the transmitter is shown by waveform Fig. 3C. Note the presence of back-scatter, noise, and the echo pulses.

Because echoes from near objects are of greater amplitude than echoes from distant objects, rough surfaces and very small objects close to the loudspeaker and microphone tend to cause unwanted back-scattering of sound, and there is also direct pick-up of the unreflected burst from the loudspeaker.

Ideally, the receiver response should be proportional to distance; low for near objects and high for distant objects, so that an echo of the same amplitude is obtained from a given object at any distance with the range of the equipment. In practice, however, it is sufficient to ensure that the receiver is pulsed off while the acoustic burst is travelling over the first few feet of its journey, and then gain can be allowed to rise quickly to maximum for the reception of echoes.

ELIMINATING BACK-SCATTER

Controlled by the gain delay pulse derived from the pulse generator, the tuned amplifier gain characteristic takes the form shown, on an exaggerated time-distance scale, in Fig. 3D. Gain is almost zero at the time of the acoustic burst, and remains low over the region of back-scatter, then rises quickly to a stable level for the remainder of the period.

The effect of gain delay is demonstrated by comparing the resulting tuned amplifier output Fig. 3E with the amplifier input Fig. 3C. Back-scatter has been virtually eliminated from the waveform, without influencing the echo pulses. Although the tuned amplifier is switched off during the transmitter burst, it is almost impossible to remove evidence of the burst from the amplifier output (Fig. 3E) because of electrical interaction between transmitter and receiver circuits, despite careful screening. However, the decoder circuit takes this into account.

The function of the decoder, which will be examined in greater detail later, is to measure the time interval between the start of the control pulse and reception of the first significant echo, while ignoring as far as possible noise, and secondary echoes.

Distance is computed by the decoder for display by a meter, and an audio bleep warning oscillator is also triggered to give an output from a crystal insert loudspeaker.

TRANSMITTER CIRCUIT

The transmitter circuit of Fig. 4 was designed to feed a peak pulsed power of more than 20 watts, at 10kHz, into a 3 ohms nominal impedance loudspeaker (12 ohms at 10kHz), using as few components as possible.

Looking first at the pulse generator of Fig. 4, C1 is positively charged through VR1, the unijunction transistor TR1 triggers into conduction, whereupon C1 discharges rapidly through R2 to produce a steep-sided positive pulse. The interval between pulses is set by adjustment of VR1, and the duration of the pulse is determined by the fixed values of C1 and R2.

COMPONENTS . . .

SONIC TRANSMITTER

Resis	tors		
RI	560Ω	R5	4·7kΩ
R2	180Ω	R6	$4.7k\Omega$
R3	330Ω	R7	330Ω
R4	220Ω		

All 10%, $\frac{1}{2}$ watt carbon

Capacitors

C1 4μF electrolytic 50V

C2 0.01 μF miniature polyester 250V
 C3 0.01 μF miniature polyester 250V

C4 4µF electrolytic 50V

C5* 4µF metallised 150V (see text)

C6 1,000μF electrolytic 50V

Transistors

TRI 2N2646 TR2, TR3, TR4 2S512 or BSY27 (3 off) TR5 AC157 or AC127

TR6 BCY39 or 2N2904

TR7 OC28

Potentiometers

VRI $100k\Omega$ horizontal preset VR2 $4.7k\Omega$ horizontal preset

Inductor

LI 2.5mH RF choke

Transformer

TI LA5 pot core (see text)

Loudspeaker

LSI 3 ohm 5in dia. (see text)

Socket

SKI Coaxial socket

Plug

PEI Coaxial plug

Miscellaneous

0-lin matrix Veroboard 4-lin \times 1-7in 3in \times 1-8in 16 s.w.g. sheet aluminium TO3 transistor mica washer and spacers 20 s.w.g. enamelled copper wire

Timing components R5, R6, C2 and C3, in Fig. 4, will cause the multivibrator circuit TR2 and TR3 to oscillate close to 10kHz, but a fine frequency adjustment is provided by VR2 (tune transmitter).

The multivibrator will only oscillate when the emitters of TR2 and TR3 are connected to the negative, rail by transistor switch TR4. In turn, TR4 is only switched on for a period of 2ms when it receives the pulse from TR1, via R4.

Driven by the multivibrator output, the complementary pair TR5 and TR6 alternately clamp C4 to positive and negative supply rails, and owing to the presence of tuned circuits L1, C4, and L3, C5, the switching waveform is converted into a sine wave by output transistor TR7.

Approximately 15V r.m.s. is developed across the 12 ohm speaker load during the 2ms pulse period.

Transformer T1 is hand wound on a Vinkor assembly, and one end of the secondary winding L3 is taken to a separate earth connection to accommodate positive or negative earth battery supplies.

REGENERATIVE RECEIVER

Receiver tuned amplifier and decoder circuits will be considered separately. Fig. 5 shows the tuned amplifier circuit

Looking first at the very important stage TR9 in Fig. 5, it can be seen that this closely resembles a regenerative tuned stage in a t.r.f. radio receiver. Regeneration does away with the need for multiple tuned stages and simplifies tuning adjustments, while still giving reasonable selectivity.

Winding L6 is phased with L7 to give positive feedback between collector and base of TR9, and thus multiplies the Q value of tuned circuit L7, C12, and TC1. Regeneration is smoothly controlled by the VR5 setting

TR9 is operated at a low collector current, near the point where gain falls off as collector current is decreased. VR4 sets the working point by establishing the amount of positive bias on the base of TR9.

A positive going pulse from the pulse generator is converted by C9, D1, and D2 into a negative pulse which counteracts the bias on the base of TR9, thus reducing gain. Owing to the slow discharge of C9 through VR4, R11, and R12 at the termination of the pulse, TR9 remains off for a period much longer than the time of the control pulse. The gain delay period is adjustable by VR4, and back-scatter control VR3.

Transformer T3 is hand-wound on an LA2107 pot

core assembly.

EMITTER FOLLOWER

The signal from the microphone could be fed straight to the base of TR9, but this would demand a microphone transformer of non-standard impedance. Also, a long run of microphone cable could cause detuning. Therefore, emitter follower TR8 is included to serve as an impedance converter and buffer stage. Transformer T2 is of the type found in hand-held, moving coil stick microphones, and has an output impedance of 50 kilohms.

As the collector load of TR9 is a high Q tuned circuit; it will present a high impedance, with large voltage amplification, to a signal of 10kHz, but side frequencies will see a low impedance and are hardly amplified at all. Overall rejection of the Fig. 5 circuit is better than 30dB for signals 1kHz above and below the centre frequency.

TUNED AMPLIFIER

Triple stage amplifier TR10, TR11, and TR12, is designed to offer some selectivity, but not enough to cause problems with instability. Emitter and coupling capacitors cause some attenuation of frequencies below 10kHz, and the response is sharpened still further by tuned collector load C16, and L8.

When viewed on an oscilloscope, the tuned amplifier output will appear similar to Fig. 3E when echo signals are present. If the oscilloscope is d.c. coupled, and can offer linear timebase sweeps at around 0·2s, then distance can be read off the X axis to give the precise location of all echoes.

DECODER

Having obtained satisfactory echo signals from the receiver tuned amplifier, these must be converted into regular shaped pulses for timing purposes.

In the decoder circuit of Fig. 6, the monostable (TR-13 and TR14), responds only to signals above a certain amplitude, determined by the VR7 setting, and thus rejects noise and small, spurious echoes.

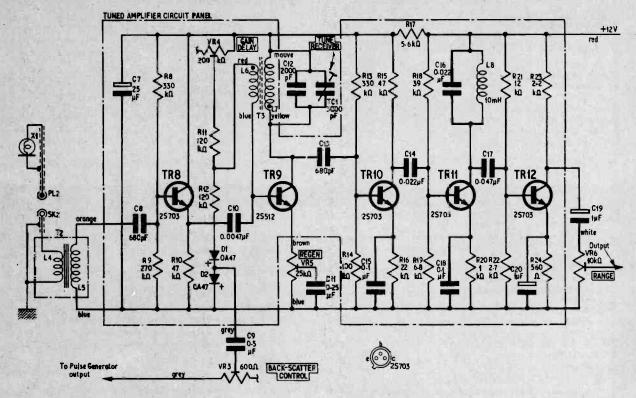


Fig. 5. Circuit diagram of the receiver tuned amplifier

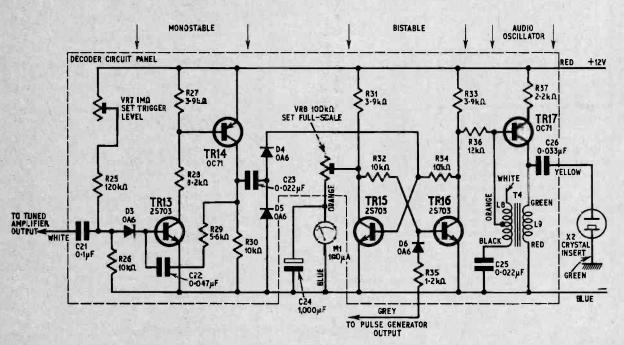


Fig. 6. Circuit diagram of decoder which converts the output from the tuned amplifier into regular shaped pulses for timing purposes

COMPONENTS . . .

RECEIVER DECODER

Re	siste	ors		
F	125	120kΩ	R32	10kΩ
F	126	I0kΩ	R33	3.9kΩ
F	.27	3-9kΩ	R34	IOkΩ
F	.28	8·2kΩ	R35	I-2kΩ
	129	5-6k Ω	R36	$12k\Omega$
	30	l0kΩ	R37	2.2kΩ
	31	3-9kΩ		

All 10%, 1 watt carbon

Capacitors

C21 0-1 µF moulded polyester 250V C22 0.047µF moulded polyester 250V C23 0.022 F miniature polyester 250V C24 1,000 µF electrolytic 12V C25 0.022μF miniature polyester 250V C26 0.033μF miniature polyester 250V Transformer

T4 LT44 transistor driver transformer (Henry's Radio)

ME 0-100µA MR-65P

Loudspeaker

X2 lin-lin diameter crystal microphone insert

TRI3, TRI5, TRI6 25703 or 2N929 TRI4, TRI7 OC71

Diodes

D3, D4, D5, D6 OA6

Miscellaneous

0-ljn matrix Veroboard 3-9in × 1-7in

COMPONENTS ...

RECEIVER TUNED AMPLIFIER

- 4				
Resiste	ors			
RB	330kΩ	R17	5-6kΩ	
R9	270kΩ	R18	39kΩ	
R10	47kΩ	R19	6.8k \O	
RII	120kΩ	R20	$lk\Omega$	
RI2	120kΩ	R21	$12k\Omega$	
RI3	330kΩ	R22	2.7kΩ	
RI4	100kΩ	R23	2·2kΩ	
R15	47kΩ	R24	560Ω	
R16	22kΩ			
A SI I	00/ Lunes carbon			

Ail 10%, $\frac{1}{2}$ watt carbon

Potentiometers

VR3 600 Ω panel mounting VR4 200k Ω sub-min horizontal mounting VR5 $25k\Omega$ panel mounting VR6 $10k\Omega$ panel mounting

All presets

Capacitors C7 25µl 25μF electrolytic 50V 680pF polystyrene 125V 0-5μF metallised 250V C9 0.5µF metallised 250V C10 0.0047µF miniature polyester 500V C11 0.25µF metallised 250V C12 2,000pF polystyrene 125V
C13 680pF polystyrene 125V
C14 0-022μF miniature polyester 250V C15 0 I µF moulded polyester 250V C16 0.022µF miniature polyester 250V C17 0.047 F miniature polyester 250V

CI8 0-1 µF moulded polyester 250V IμF miniature electrolytic 15V C20 IµF miniature electrolytic I5V Trimmer Capacitor TCI 3,000pF Bulgin C.P.7

Inductor

L8 10mH RF choke

Transformers

T2 50kΩ moving coil microphone, matching transformer

T3 LA5 pot core (see text)

Transistors

TR8, TR10, TR11, TR12 2S703 or 2N929 TR9 2S512 or BSY27

Diodes

DI, D2 OA47

Plug PLI Coaxial plug

Socket

SK2 Coaxial socket

Microphone XI DM107 or similar

Miscellaneous

0-lin matrix Verobcard 6-9in × 1-6in 40 s.w.g. enamelled copper wire

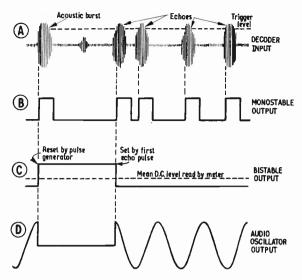


Fig. 7. Waveform sequence appearing at decoder. (A) Typical input signal. (B) Response of monostable. (C) Output of bistable. (D) Audio oscillator output

Demodulated by D3, the leading edge of a signal envelope triggers TR13 on, which in turn switches TR14 on. With the collector of TR14 now close to the positive rail, TR13 remains biased on by the slow charge of C22 through R29, giving a rectangular pulse output from the collector of TR14, of length dependent on C22 and R29 values.

Owing to the positive voltage present on the base of TR13, D3 is reversed biased, and blocks further signals until the monostable switches back into its stable state. The duration of the monostable output pulse determines the minimum distance which can be measured. If C22 is made smaller in value, a point will be reached where the monostable begins to deliver two pulses from each input signal.

BISTABLE

Looking next at the bistable in Fig. 6; a pulse via R35 and D6 from the pulse generator, at the commencement of each acoustic burst, will switch TR15 off and TR16 on. With the collector of TR15 almost at positive rail voltage, C24 will begin to charge via R31 and VR8. If no echo signals are present, C24 is allowed to become fully charged and the meter M1 will read full scale.

In the event of an echo being repeatedly received some time after each transmitter burst, however, the leading edge of a pulse from the monostable will switch TR15 on and TR16 off, thus interrupting the slow charge of C24, and giving a meter reading which is directly proportional to the distance of the nearest echo producing obstacle.

DECODER WAVEFORMS

The decoder waveforms of Fig. 7 show clearly the sequence of events. Fig. 7A is a typical input signal to the decoder, consisting of an acoustic burst waveform followed by four sizeable echo waveforms spaced in time. Fig. 7B gives the response of the monostable to the Fig. 7A signals, with a pulse for each signal envelope.

As the acoustic burst from the monostable, and the reset pulse from the pulse generator occur simultaneously, the bistable ignores the former, but will be

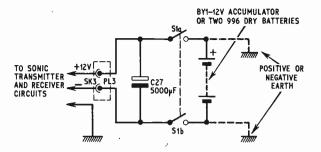


Fig. 8. Power supply circuit suitable for car battery or dry battery working. Battery impedance is minimised by large electrolytic

COMPONENTS . . .

POWER SUPPLY AND BOX Capacitor C27 5,000μF 15V

BATT 12V accumulator or two 996 dry batteries

Plug PL3 Non-reversible 2 way

Switch
Sl Toggle d.p.s.t.

Socket SK3 Non-reversible 2 way

Miscellaneous

Universal chassis 6 in \times 8 in \times 3 in with extra 6 in \times 8 in top plate and 8 in \times 3 in side (Home Radio)

set by the leading edge of the first echo pulse to be received, see Fig. 7C. The bistable can only respond again after it has been reset, and gives an output of duration exactly corresponding to the time taken for the sound to complete its journey.

The inverse occurs with the audio oscillator waveform Fig. 7C; in the event of no echo there is no output from TR17 but an echo from a close obstacle gives an audio oscillator output of long duration.

POWER SUPPLY

There are two possible snags when the Obstacle Locator is powered from a car battery; interference and voltage fluctuations.

Charging circuit contacts, dynamo brushes, and agnition spark gaps generate quite a lot of interference, which could increase the sonic receiver noise input to the point where maximum range suffers.

Although the situation could be improved by incorporating filters and some form of voltage stabilisation, the easiest and cheapest solution is to run the equipment from suitable dry batteries.

Average current consumption of the Obstacle Locator is in the region of 50mA, so two 996 batteries coupled in series will give 100 hours operation.

A power supply circuit suitable for car battery or dry battery working is shown in Fig. 8. The large electrolytic C27 helps to minimise battery impedance.

Part two next month will cover the construction, testing, and calibration of the Sonic Obstacle Locator.

SIREN



Most readers are probably familiar with the banshee wail of a siren. Rising from a low frequency beginning, the sound rises up in a smooth glissando through baritone to soprano, hovers, then falls, this being maintained through a series of undulating peaks.

For novelty appeal, this particular offering is pitched approximately to those sounds made by American police cars, albeit very much attenuated in decibels. For this reason alone it should prove an attractive project for youngsters.

PULSE GENERATOR

The active ingredients in this design are an *npn* and *pnp* complementary transistor pair arranged to form an astable multivibrator, which would be free running if R2 was connected directly to the positive rail of Fig. 1.

With such a connection the frequency would stabilise around 4kHz, this rate being determined by the time constant, or CR product, of R2 and C2. Mathematically, the pulse frequency is said to vary inversely as this product, which simply means that an increase in value of either R2 or C2 means a decrease in frequency, or alternatively with the values decreased the frequency rises.

The conduction cycle commences with the charging of C2 via R2. As discussed last month, the charging process takes the form of an exponential growth curve of voltage. No conduction takes place in the transistor pair until C2 reaches about 0.7V.

CUMULATIVE ACTION

At this voltage TR1 moves into action, passing current to its load TR2. Up to this point the collector of this transistor is at -9V relative to the emitter and with conduction this goes positive. This is immediately reflected to the base of TR1 which accelerates the overall switching of the transistor pair.

This self-regenerative feedback process, or cumulative action, results in a very sharp pulse at the speaker.

RISE AND FALL

In describing the action of the constant frequency generator it was arranged that R2 was fed from the positive line.

To produce the frequency variations necessary for the sireh sound, R2 is in fact connected to the mid-point of a timing circuit C1, R1. S1 is introduced into this net so that rise and fall periods are arbitrary.

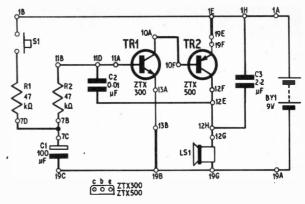
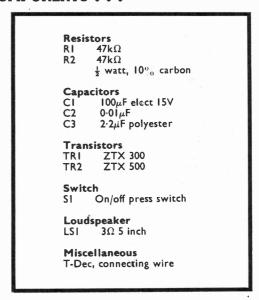


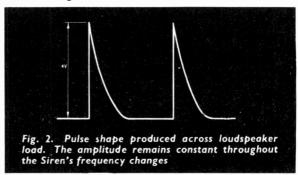
Fig. 1. Circuit diagram of the Siren; link wires are shown as thick lines. Note that an electrolytic capacitor should not be substituted for C3

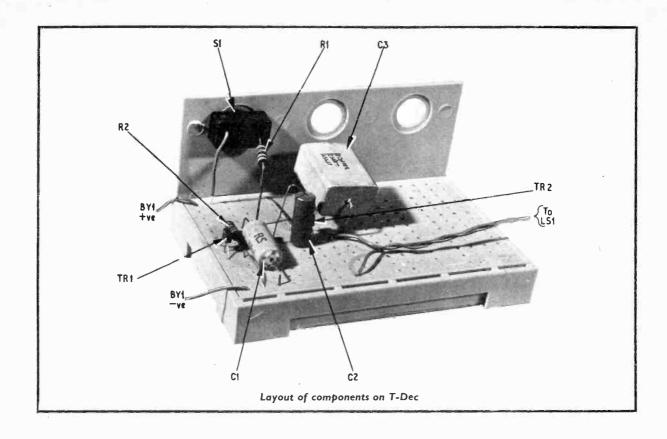
COMPONENTS . . .



With S1 closed, C1 charges on a relatively long time constant to a maximum of 4V. A rising voltage across R2 and C2 means, in effect, a diminishing time constant, so the overall frequency starts to rise.

Just as the charge transition is smooth on C2, so is the frequency change which starts at zero and glides up to about 2kHz. The output at LS1 remains constant at 4V during this period, the waveshape produced being shown in Fig. 2.





The release of S1 causes the frequency to fall. Since C1 is charged to about half the rail volts, the discharge period is reduced by this factor.

INCREASING THE OUTPUT

Without the inclusion of C3 the power available at the output would be negligible. This capacitor has the effect of increasing the pulse width and so raising the mean d.c. level.

It is most important that a capacitor with a low

power factor be used here; an electrolytic type just will not do.

To alter the frequency of the siren C2 should be changed. An increase in capacitance will mean a reduction in peak frequency and a decrease will raise it.

ALTERNATIVE CONSTRUCTION

A permanent assembly of this unit can be made on Veroboard. Component mounting can be identical to that given for the T-Dec as the hole matrix for the board is identical.

NEWS BRIEFS



Scouts Communications

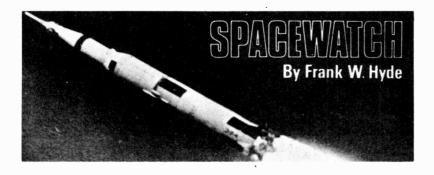
THE CENTRE of the 3rd Bracknell Scouts was turned into a complete communications centre for a weekend in October by equipment loans from the nearby Racal Electronics Group. The Racal Amateur Radio Club GB3RAC, also loaned the Scouts equipment and the members assisted operations for day and night watches.

The centre had four complete transmitting/receiving stations, five separate monitoring stations manned by Scouts, a news and weather bureau with all necessary sound and TV monitoring receivers, teleprinter, and Mufax picture receiver.

Pictures of weather charts included one from Khabarovsk in Eastern Siberia. Other weather informa-tion was intercepted from weather ships in the Atlantic.

Jamboree contacts were made with numerous Scout Groups throughout the world, including Australia, S. Africa, U.S.A., Middle East and Europe.

Over 200 Scout Leaders and Scouts shared watch duties throughout the weekend and there were over 200 spectators. The photograph left shows part of the communications centre operating.



A DECADE AHEAD

As 1971 opens it is perhaps worthwhile to look at the next decade as planned at this moment. Certain programmes have already been put in hand and these have been covered by previous *Spacewatch* articles. However, in order to take an overall look at the next decade some of the programmes will of necessity be noted again.

One of the most important items in the programme is to find ways of reducing the cost of space operations. For this reason the development of the space shuttle will be a priority venture. The prospect of a re-usable shuttle vehicle which can travel between the Earth and space stations in Earth orbit offers a system that is within the range of present technology. Already the design stage has reached a definitive study. The vehicle in its present form will be launched by a booster from a rocket pad and return to Earth to land like an aircraft.

SPACE TUGS

To this must be added the rather newer proposal for a space tug which would serve for goods and personnel / transport from space station to space station. Other operations to be undertaken would be the raising of satellites from lower orbits to higher and synchronous orbits and the firing of automated probes from Earth orbit into orbits that will enable a further investigation of the planets. The tug would ride in the shuttle cargo compartment. Already these programmes envisage international participation in a greatly extended manner.

In the past decade there has been considerable co-operation so far as contractors were concerned, but the new thinking is more in the terms of a cost sharing as well as contracting between the U.S.A., the United Kingdom, Federal Republic of Germany, France, Belgium and Italy. This could mean that the European Nations will provide half of the cost of the space tug (about 1,000 million dollars). The cost of the space programme for the shuttle is estimated at about 4,500 million dollars, and for the 12 man space station

upwards of 4,000 million dollars. There is to be a standardisation of technique, and both vehicles should be operative by 1978.

SKYLAB

This will be a large orbiting workshop which will examine the capability of men to live and work in space over long periods. It will use systems already developed for the *Apollo* missions. Later, there will be even more sophisticated vehicles with the exchange of personnel on varying tours of duty.

EARTH RESOURCES SATELLITES

One of the major projects already well under way is that of surveying the earth to assess its potential for the future. This programme will have an unprecedented facility for the study of crops, the location of new sources of water and the prospecting for mineral deposits. There will be the more extended use of space vehicles in the fields of weather forecasting, communications, air traffic control, geodosy map making and navigation. There will also be concentration on the use of television for education particularly directed under-developed toward the

The Apollo programme, although curtailed somewhat from the original one, will nevertheless offer an exciting new extension of moon exploration. Apollo 14 site will be the same as that which was originally scheduled for Apollo 13, that is the Fra Mauro rugged region. The site for Apollo 15 has now been decided; it will land in the lunar plain much further North than any previous mission. There is an area known as the Hadley-Apennine region which has the 8,000 ft. Apennine range on one side and the 60 mille long meandering canyon, 600 ft. deep, known as the Hadley Rille, on the other. It is an important region and will yield a great deal of valuable scientific data.

The astronauts will be able to gather material from the base of the mountains, some of which may date from the old lunar crust, which existed more than 4,000 million years ago. It is thought this may be older

than the Mare Imbrium basin which was probably formed by meteoritic impact. There may also be materials, thrown out by this cataclysmic impact, which originated deep within the moon.

On this mission the Lunar Rover vehicle, electrically powered, will be in operation for the first time. The crew will be able to go some miles from the base of operations in order to carry out experiments, and set up scientific instruments. The crew will be David Scott, Alfred Warden and James Irwin. Launch date is likely to be July 1971. The following two Apollo missions (16 and 17) are scheduled for January 1972 and June 1972; no sites or crews have been decided at the moment.

MARS PROBES

These have already been noted in some detail in *Spacewatch* for June 1970 and this programme gets under way with the first launches in 1971.

GRAND TOURS

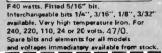
These will take advantage of the position of the planets in the solar system during the next few years. In the case of the first mission the next possible time for this to be done will not come again for 200 years and the second planned mission could not be undertaken for another 100 years. Therefore, these programmes are of great importance, for to achieve the objectives of these missions with our present technology, the gravitational effects of the planets will be needed for an effective gain in momentum for the vehicles as they move from one area to another. These will of course be unmanned. The manned tours will need nuclear or some other source of power for this to be successful. The testing of nuclear powered rockets will be made in 1978 with unmanned craft.

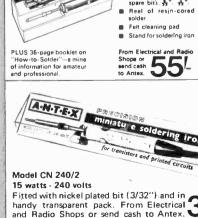
The Mariner Venus-Mercury tour will begin the series and for the first time man may be able to see a close-up of the surface of Mercury. The next will begin in 1977 with one or perhaps two unmanned craft launched toward Jupiter.

When the spacecraft arrives in the vicinity of Jupiter it will swing round the planet under the influence of the local gravitational field; four years later it will reach the planet Uranus where again gravity will influence the path to project it toward the end of its journey five years later at the planet Neptune in 1988 after a 3,000 million mile trip.

A further launch will be in 1979. The route will on this mission be more direct to Jupiter and by 1980 will swing round Saturn and will head for Pluto which it should reach five years later. Pluto itself is the most distant planet from the Earth and will at that time be some 3,700 million miles distant.







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PART SEVEN—By R. W. COLES TTL FLIP-FLOPS

THE range of bistables, or flip-flops as they are usually termed in i.c. parlance, is more varied in the TTL family than in either DTL or RTL. There is, in fact, a flip-flop to suit every occasion, or for those who dislike using a different component for each specialised application, there is one flip-flop which may be used for any of the normal storage circuits.

Flip-flops are one of the most important elements in a logic system, all types having one feature in common. They provide storage of one binary digit for any length of time which may be necessary, as long as the power is

applied.

The differences between the various types occur only in the way that the data to be stored is "written" into them, their outputs always consist of a true and an untrue indication of the digit they contain. By connecting them together in the appropriate fashion, it is possible to build many types of counters or dividers, shift registers, and memories, which may contain a complete binary word.

DATA ENTRY

Data may be entered into a flip-flop either asynchronously, i.e. at any time, regardless of the original contents or the state of the clock pulse, or synchronously, i.e. the data is entered only when a clock pulse "enables"

the data inputs.

The flip-flops in the TTL family usually have both synchronous and asynchronous inputs, which makes them most versatile in use. In fact, so many useful features have been packed into such a small package, that the discrete, specialised designs of ten years ago seem primitive in comparison.

Apart from the usual types of data entry mentioned above, TTL synchronous inputs can be further subdivided into either those which employ the positive edge of the clock pulse for gating (edge-triggered), or those which employ both edges of the clock pulse

(master/slave).

EDGE TRIGGERED TYPE

With the edge-triggered type, assuming that the data to be entered is waiting at the inputs, as the clock pulse starts to rise this data is entered into the storage latch, and is thus immediately present at the outputs. As soon as the clock pulse has passed the threshold of the input gates, the input data is locked out, and can have no effect on the stored information until the clock rises again.

MASTER/SLAVE

The master/slave type is a little more complex in operation, as the individual flip-flops of this type really consist of two interconnected bistable sections, termed the master section and the slave section. Information enters the master section as the clock pulse rises, but this information does not appear at the outputs immediately because of the interposition of the slave section.

The slave bistable is also clocked, but through an inverter. The slave sees a positive clock pulse when the main clock pulse is actually falling, and enters the data present at the output of the master latch, and transfers it immediately to its outputs, which are also the outputs of the flip-flop as a whole.

Ignoring what goes on inside such a flip-flop and looking at it just as a "black box", information enters on the positive excursion of the clock, and appears at the outputs on the negative excursion. A simplified logic diagram of this type is shown in Fig. 7.1, together with a diagram of clock action.

Master/slave JK flip-flops form the backbone of TTL counting circuits, giving maximum versatility at the expense of a slightly higher power dissipation and slightly lower maximum toggle frequency than the edge-

triggered type.

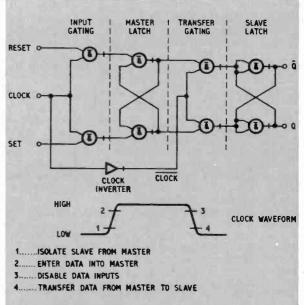


Fig. 7.1. Master/slave flip-flop. This is a simplified explanatory arrangement, using the master/slave principle. As it stands, it is a set/reset type, but JK operation is simply achieved by connecting the Q output to the reset input gate, and the \overline{Q} output to the set input gate. No preset or clear inputs are shown

A few examples of the many flip-flops available of both edge-triggered and master/slave type are given with their package outlines in Fig. 7.2. Note that some JK types have several J and K inputs, so that gated inputs for synchronous counters will not require separate gate packages.

Some of the flip-flops have no preset input—only clear. These are useful for counter circuits, where the asynchronous inputs are not needed other than to set up the all zeros condition before a count sequence.

TYPE-D FLIP-FLOP

There is one "odd-ball" which is a little different from the rest—the 7474 type "D" flip-flop. This one is intended specifically for shift-register applications, and

belongs to the edge-triggered group.

It embodies full asynchronous input capability, but has only one data input D. The other data input is obtained internally by inverting the information on the D input. This means that only one connection is necessary to transfer data from one stage of a shift register to the next.

Perhaps this is not so obvious an advantage until you consider that many shift registers require inputs from several sources, in which case the external gating is cut

by half when using the type D.

Although it is intended only for register use, it may be used in ripple counters by connecting the \overline{Q} output back to the D input, but this is by no means the best choice of flip-flop for such use.

TTL COUNTERS

Having discussed what types of flip-flop are available, it is possible to have a look at some applications. To start, let us examine counting circuits. Simple ripple counters were described in the sections on RTL and DTL, and should help here to understand the principles involved in TTL.

RIPPLE COUNTER

Ripple counters are the simplest type of design using TTL i.c.s; all that is necessary is to connect the Q output of the first flip-flop to the clock input of the second, and so on, the input being connected to the clock of the first flip-flop in the chain.

The maximum number of separate states is given by 2^n , where n is the number of flip-flops so connected. Thus a four-stage counter has sixteen states, and a

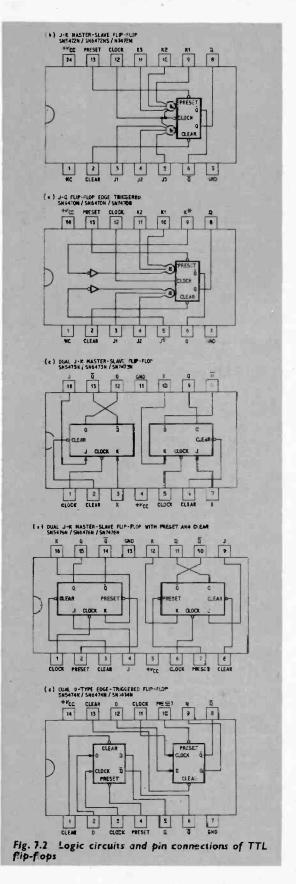
five-stage counter has thirty-two states.

If a count which is not a binary multiple is required, it is necessary to force the counter to skip some of its usual states. There are several ways to achieve this, the simplest method being to detect the desired final count +1 with a NAND gate, and use its output to reset all the flip-flops in the counter to zero.

This method may, however, result in a "race" condition if one of the counter stages resets more readily than the others, causing the NAND gate output to disappear before all stages are reset. For this reason it is not commonly used, although it is possible to include some sort of delay, such as a monostable, in the reset path to ensure that the reset pulse will last long enough to do its job properly.

IMPROVED DIVIDER

A much better method of dividing by a number which is not a binary multiple is to feed the flip-flop outputs which are in the 1 state at the desired final count, along with the counter input, or clock, to a NAND gate, and use the resultant output to set all the counter stages to 1.



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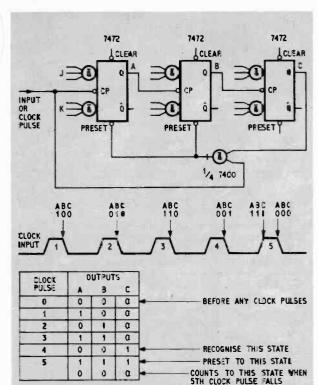
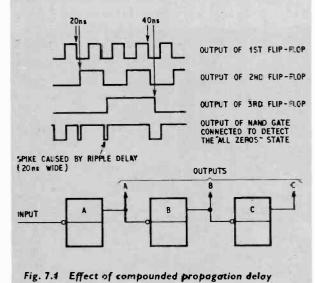


Fig. 7.3. Improved divider with truth table



When the final clock pulse falls (assuming JK master/slave flip-flops), the counter will count to the all The appearzeros state, ready to begin another cycle. ance of the all 1 state during the final clock pulse is unlikely to cause a problem, although this depends on the application.

To see this system operating, take as an example a divide-by-five counter using type 7472 JK flip-flops. The nearest power of 2 which is higher than 5 is 23, so a three stage counter will be necessary as a basis of the

design.

The final counter state will be binary four, 100, because the first state of the counter will be zero, i.e. 000. The only flip-flop which has an output of 1 at this stage is the third one, so we must feed this output together with the clock into a 2-input NAND gate, and connect the output of this gate to the preset inputs of the first two flip-flops.

The final circuit is shown in Fig. 7.3, together with a truth table and the clock pulse action. Note that the J and K inputs are not used in this application, and are therefore left disconnected, simulating a 1 input. They can, in practice, be connected to $+V_{cc}$ to ensure this state.

PROPAGATION DELAY

The counters we have considered so far have all been ripple types, and, as such, they all suffer from a drawback which can be serious in some applications.

If we look closely at such a counter which is in the all 1 state, it will be required to count to the all zeros state on the arrival of the next clock pulse. The final flip-flop in the chain will not be able to do this until all the stages preceding it have already done so; after all, it gets its own clock edge from the flip-flop immediately before it, which in turn gets its clock edge from the stage before it, and so on.

This all sounds quite straightforward, until we consider the propagation delay incurred from each stage, which in the case of the 7472, is 20ns. This seemingly insignificant delay is multiplied by the number of flip-flops there are in the counter, and may reach alarming proportions in a counter working at high speed and with a long cycle length.

Even so, it need not be a problem unless several of the counter outputs are gated together. Spikes appearing at the gate output at the wrong time may cause other circuits connected to the gate output to mistrigger. To illustrate the effect of this ripple delay, Fig. 7.4 shows the output waveforms obtained from a three stage counter, with the error due to the delay magnified.

To overcome this ripple delay error, a different kind of counter called a synchronous counter is used. With this type, the clock input is fed to every stage, not just the first. Input gating is used to determine the conditions when a flip-flop is required to "toggle", or change state.

Next month's article will continue with synchronous counters, shift registers, and ring counters using TTL flip-flops.

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WASH-WEE

An electronic device for intermittent wiping and automatic wiper operation with washers, on cars with 12 volt field coil wiper motors.

A FACILITY not provided as standard on older cars but which proves to be very useful in the British climate is that of intermittent operation of windscreen wipers. Another useful facility is windscreen wiping initiated by operation of the windscreen washer control and continuing for a limited period after its release. Both facilities are provided by the circuit to be described which can be added to the existing circuits in the car without impairing the normal operation of the standard controls.

The circuit takes advantage of the windscreen self-parking mechanism fitted to most cars; reference to back numbers of practical motoring magazines will show that this mechanism can be fitted to almost any vehicle very easily at little cost. The circuit is also basically designed for negative earth connection but a modified version to suit positive earth connection is given. The unit cannot be operated with the latest permanent magnet wiper motors, or in conjunction with 6 volt systems.

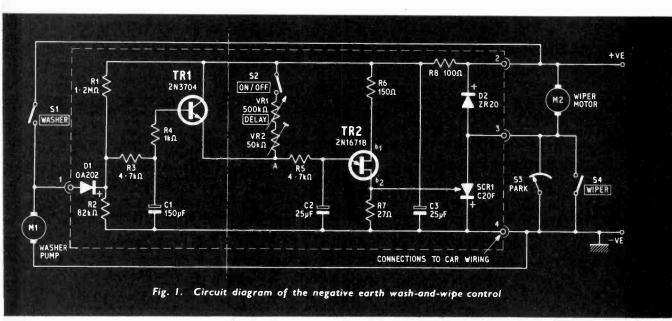
CIRCUIT DESCRIPTION

The circuit diagram for negative earth connection is given in Fig. 1 and the mode of operation is as follows.

The thyristor SCR1 is connected in parallel with the manual switch controlling the windscreen wipers (S4) and when triggered will initiate wiper action. A small amount of wiper action closes the self parking switch S3, short-circuiting SCR1 and turning it off; the wipers continue under the control of the self parking switch to the end of a single sweep. If SCR1 is not in a triggered state at this time, no further wiping action will take place until SCR1 is again triggered or the wiper switch (S4) is closed.

The function of diode D2 is to bypass the inductive surge current generated when the motor is switched off. The necessary trigger pulses for SCR1 are provided by the unijunction oscillator TR2 at intervals determined by the time constant (VR1 + VR2 + R5)C2. The function of R5 will be described later. VR2 is a preset potentiometer which determines the fastest intermittent rate and VR1 is the rate control which can be omitted if fixed rate operation only is required, in which case VR2 determines the rate. Switch S2 is the on-off control and is incorporated in VR1.

With the component values given, the sweep rate can be adjusted to cover a range from one sweep every 1.5 seconds to one sweep every 1.5 seconds. Resistor



CONTROL

By D.E. VAUGHAN

R8 and capacitor C3 form a filter which prevents voltage disturbances on the battery line from producing false trigger pulses. If R5 is omitted together with that part of the circuit to the left of the chain dotted line (the components on tagboard A), the circuit as described will provide the intermittent sweep facility alone.

LIMITED CONTINUOUS OPERATION

The limited continuous wiper operation is obtained in the following way. In the quiescent state, with S1 open, point A (Fig. 1) is approximately at the same potential as the junction of R1 and R2 and, since this is less than ηV_b , the unijunction oscillator will not function. When S1 is closed C1 is charged rapidly through D1 and R3 and by emitter follower action in TR1 point A is raised above ηV_b , allowing the unijunction oscillator to function at a rate determined by the time constant $R_s C_2$. (Where ηV_b is the intrinsic stand-off ratio.)

If the resistor R5 has the minimum value consistent with reliable operation, a high repetition rate is obtained so that there is minimal delay at the end of each wiping stroke before SCR1 is re-triggered. When S1 is opened C1 discharges slowly through R3 in series with the parallel combination of R1 and R2 towards a potential $R_2V_{10}/(R_1 + R_2)$ and the potential of point A follows. When this falls below V_b the unijunction oscillator again ceases to function.

The effective time constant of the component values suggested allows three or four sweeps after the opening of S1, depending upon the speed of the windscreen wiper motor. The elaborate charge-discharge circuit

is incorporated to prevent an unwanted sweep occuring when power is first applied. It will be noted that the limited continuous wiping facility is obtained irrespective of the state of S2. If the latter is closed, intermittent operation automatically follows the end of the continuous wiping period.

COMPONENTS

The modified circuit for use in vehicles with positive earth connection is shown in Fig. 2. Two additional components are required to provide the necessary improvement in false trigger pulse suppression.

The choice of *npn* transistor TR1 and unijunction transistor TR2 are in no way critical, the particular types listed being those which were at hand. The thyristor SCR1 should have current rating at least equal to the running current of the windscreen wiper motor, typically 3 to 4 amps, although in the interests of reliability the thyristor should be of sufficiently high rating to handle the hot stall current, typically 7 to 8 amps.

Diode D2 should have a forward current rating equal to at least half of the normal running current. Other component values are selected to suit the particular devices used and to provide the desired timing intervals.

CONSTRUCTIONAL DETAILS

The simplicity of the circuit means that any form of construction which is convenient may be used without difficulty. However, details of the form of construction used in the prototype are given here as a guide.

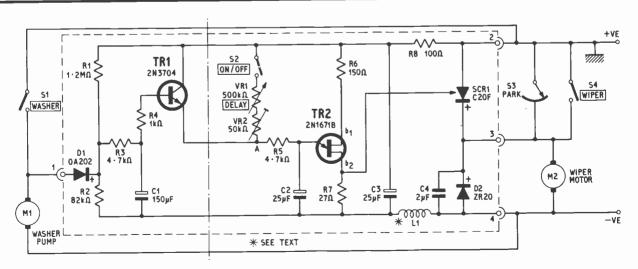


Fig. 2. Circuit diagram of the positive earth wash-and-wipe control

GATE CATHODE 62 0 2N1671 ANODE SCR1 C20F C ATHODE 6 0 0 0 2N3704

Fig. 3. Semiconductor connections for the components listed

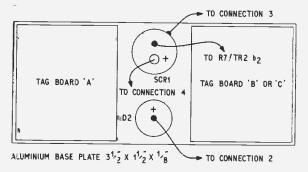


Fig. 4. General layout of the tag boards

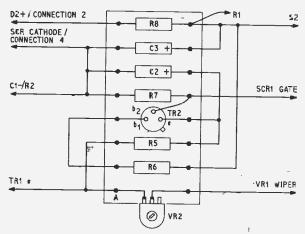


Fig. 5. Tag board B, for negative earth version

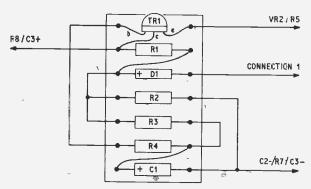
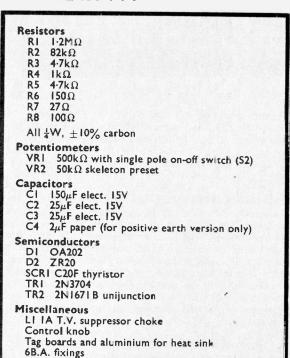


Fig. 6. Tag board A, for both versions

COMPONENTS . . .



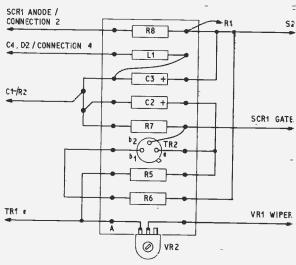


Fig. 7. Tag board C, replaces tag board B for positive earth version

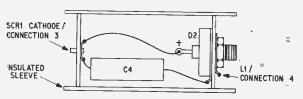


Fig. 8. Sub-assembly for positive earth version

For the benefit of readers who are not familiar with the unijunction transistor and the thyristor, details of the electrical connections are shown in Fig. 3. Although these are typical, should other types be used the relevant manufacturer's information must be followed.

NEGATIVE EARTH VERSION

The general layout for a negative earth system is shown in Fig. 4, SCR1 and D2 being fixed directly to the base plate which acts as a heat sink. Two tag boards are also attached to the base plate. Since the base plate is live to the circuit, it is enclosed in a small plastics box using nylon screws as mounting studs; this box can be fixed at a convenient point behind the car facia panel (the base plate must not be earthed to the car).

Detailed component layouts on the two tag boards are shown in Figs. 5 and 6. Although VR2 is shown on the circuit diagrams as a preset component, once its value has been found it can be replaced by a fixed resistor.

Two groups of leads are taken from the box, one group going to S1, S2 and VR1 mounted on the facia panel and the second passing through the bulkhead to the washer pump, the wiper motor and to the positive and negative battery connections.

POSITIVE EARTH VERSION

Because the stud connections to SCR1 and D2 are not common when a positive earth system is required, a slightly different form of construction has to be adopted. The general layout of the base plate differs from that shown in Fig. 4 in that D2 is omitted and one of the tag boards carries an additional component (L1)—Fig. 7 replacing Fig. 5.

Diode D2, together with the additional capacitor C4, is mounted as a separate assembly in a small insulated tube (see Fig. 8) fixed near the wiper motor. In all other respects the details of construction are similar.

CONNECTION TO WIPER MOTOR

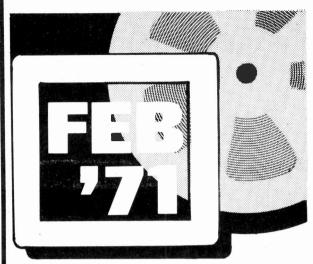
Examination of the electrical connections to a wiper motor already fitted with a self-parking switch will show two live leads and an earth return. Reference to the workshop manual, or a few minutes work with a voltmeter, will establish which of the live leads is connected to the battery and which is the switch lead. The lead from the base plate (i.e. the common connection of the thyristor and diode anodes) is connected to the switch lead terminal. It is stressed that no modification to the wiper motor is required unless the motor does not have the self-parking facility.

In the case of wiper motors not fitted with a selfparking switch, the first step is to make the necessary modifications. When the modifications have been completed the connection can be made as previously described.

WASHER PUMP SWITCH

An electric washer pump switch is sometimes fitted in cars but is normally connected between the pump and earth. If this is so it is a simple matter to reconnect is as shown in the circuit diagram (Fig. 1).

Although the control for manual and vacuum washer pumps is not electrical, it should be possible to fit a microswitch, actuated by the normal control, in place of the washer switch. Other methods of switching may well be possible and no doubt some readers will add this facility to non-electric washers.



TAPE SLIDE CONTROLLER

This electronic control system can be added to any tape recorder to control an automatic slide projector in sequence with a recorded sound track. It has been designed so that no modifications are necessary to the tape recorder or the

SUBMARINE CHASER

A follow-up "war" game to Operation Seasearch (December 70). This game uses the same chassis and much of the wiring of Operation Seasearch but introduces depth. In Submarine Chaser, "Asdic" is used to narrow the search and the destroyer must adjust his depth charges correctly to destroy the submarine.



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By C.W. Smedley C.Eng., M.I.E.E.

Southall College of Technology

APE recording has been covered by many text books, and the general principles are well known. The explanations given of bias, however, tend to be vague, and those who would like to pursue the matter are often deterred by the B/H loop, with its associated mathematics, implied if not stated. This article is intended to be a reasonably valid description avoiding the use of complicated analysis.

The object is to record signals, which can be audio, video, digital, etc. on a tape which is either coated or impregnated with magnetic material, and the system

described here will be longitudinal.

THE MAGNETIC CIRCUIT

The recording/playback head consists of a winding on a ferromagnetic core, and current fed to the head will produce a magnetic flux. Fig. I shows the relationship between flux, B, and the magnetising force H, which is proportional to ampere-turns. If the current is increased slowly from zero, the value of flux can be plotted against the value of H, and the line OA will be traced.

At some point, A on our diagram, the value of flux will remain constant for any further increase in H, and the core is said to be saturated. P represents this maximum value of flux and S shows the corresponding value of H. If the current, i.e. H, is now reduced to

zero, the flux decays along the curve AR, and R is the remanent value of the flux.

The ratio of OR to OP is called the retentivity of the core material. Some ferrites produce a curve in which OR and OP are practically equal; these are called square loop materials and are used in computer stores.

HYSTERESIS LOOP

If the current is now increased in the reverse direction. at some point the flux will have fallen to zero, shown as point C. Further increase will bring the flux to point D, which is reverse saturation, identical in value with point A, but opposite in polarity. Reduction of current, towards zero, will reduce the flux, and the curve DQ will be traced out.

Note that this curve does not pass through point O, but rises, as current is increased in the original direction, through E to A. It can be seen that, for any value of H, there are two possible values of B, and the perpendicular drawn at the point where H equals X cuts the B/H loop at points Y and Z. When H is rising, the value of flux corresponding to Y is less than on the original trace, OA, whereas, when H is falling, the value of flux corresponding to Z is higher. lagging effect is called hysteresis, and the curve ACDEA is called the hysteresis loop.

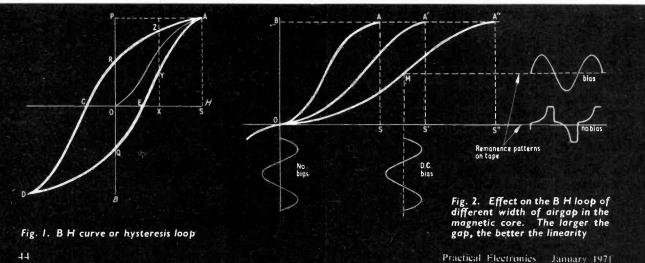
THE AIRGAP

If the magnetic core formed a complete ring there would be no flux outside the ring, and a tape drawn across it would be unaffected. To produce flux outside the ring it is necessary to break the magnetic circuit, i.e. to have an airgap, and it will be at this point that the moving tape will be brought into contact with it.

The airgap will increase the reluctance of the magnetic circuit, and the flux for any given value of head current will be less than the value without an airgap. Another effect, however, offsets this apparent loss, and Fig. 2

has been drawn to illustrate it.

If we assume that the B/H loop is very narrow, then let OA represent the loop with no airgap. OA' shows the loop when a small gap is introduced, OA" the loop for a bigger gap. The first trace, OA, shows extreme curvature near the ends, and would not, in fact, be linear at any point. OA' is much more linear, although the slope is less, and OA" is better still, although the value of H necessary is increasing with the size of the airgap. This is not too important, since the amplifier gain can be increased to compensate. For the sake of linearity, therefore, the gap should be as big as possible.



But it can be shown that, for any given tape speed, the bigger the gap the poorer is the high frequency response of the system, and the two fundamental requirements appear to be in conflict. The practical solution is to make the actual recording/playback gap as small as possible, then to insert a second gap diametrically opposite to it.

DIAMAGNETIC GAP

Although the term airgap remains in use, the actual gap must be filled, for two reasons. One, if left open it would collect magnetic material from the tape, to re-form a closed magnetic circuit. Two, the external flux would be low in value. The gap is closed by a material which is both hard, to resist wear, and diamagnetic. Brass is a good example, and is commonly used. The flux now "fringes" out across the gap to make recording more effective.

The actual distribution, however, is not at all like the pattern across the open end of a horseshoe magnet. The brass insert acts as an obstruction to the flux and the tape itself closes the magnetic circuit. The flux can be said to enter and leave the tape at right angles to the surface of it, but all of the flux does not leave the core at the edge of the gap. It is spread each side of the gap to form a fringe which tails off within a short distance, and the electrical gap is therefore greater than the physical gap in the head. It is important to remember this when considering erasure.

THE NEED FOR BIAS

The effect of recording a signal has been shown by superimposing it on the vertical axis of Fig. 2. The recorded pattern on the tape will show the curvature near to point O, and harmonic distortion has been introduced. To prevent this effect the working area should not include the extreme curvatures near zero and saturation.

Direct current bias can be used, for example sufficient current can be passed through the record head so that every tape element is, in the absence of a signal, magnetised in the same direction, to the same degree. Such a current would correspond to a value of H somewhere between S and S', so that the effective working point is M on the curve OA''. The signal to be recorded is now shifted bodily to the right, so that it becomes symmetrical about the vertical drawn through M.

This bias system closely resembles that used for valves and transistors, where the transfer, or mutual, characteristics exhibit similar curvature at cutoff and bottoming.

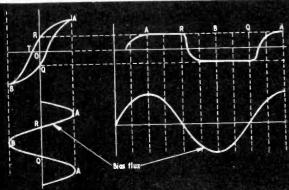


Fig. 3a (left). Applied flux on the H axis of the magnetic loop is transferred to the tape in distorted manner due to remanent magnetism, which is plotted in Fig. 3b (right) and shown against the sine wave producing it

Practical Electronics January 1971

Although this d.c. system is used on cheap tape recorders, there are two major limitations. One, the peak to peak signal amplitude must be only half the value of that when no bias is used. This can be partially overcome by shifting the d.c. bias point, but the benefits are marginal and still include some distortion. Second, every particle on the tape is magnetised and, on playback, will make a contribution to the output voltage in the form of noise, mainly high frequency (hiss). For these reasons supersonic (a.c.) bias is now almost universal.

HIGH FREQUENCY BIAS

The frequency chosen must meet two requirements: it must be high enough to be filtered out on playback without attenuating the wanted signal and, in general, should not be less than three times, preferably five times, the highest frequency to be recorded. At the same time, it must not be too high because the record head is highly inductive. The practical value lies between 50 and 75kHz and a sinewave form is essential to prevent head magnetisation by second harmonic. LC oscillators of the Hartley or Colpitt type are popular, and the stage often "doubles" as audio output on playback.

To understand the full effect of a.c. bias, two points should be kept in mind. One, the gap includes the fringe each side of it, and the actual flux entering the tape from the head is almost perpendicular to the surface. Two, the flux is alternating at bias frequency, and the actual value affecting each tape element will be the instantaneous value as that element leaves the fringe. Each tape element, or particle, is therefore exposed to a flux which can have any value from zero to maximum, of either polarity, but the resultant magnetic pattern on the tape is not sinusoidal.

Fig. 3a will show this more clearly, and the applied flux, H, is shown to have sine wave form. Point A on the sine wave corresponds to saturation, i.e. point A on the B/H curve, and any tape element exposed to this value of flux will have a remanent magnetisation equal to point R when it leaves the fringe. Following elements will be exposed to a flux decaying towards point R but all will have the same remanent value as point A.

REMANENT FLUX

Fig. 3b shows the remanent flux plotted against the sine wave producing it, and it can be seen that the waveform written on to the tape has a flat top from point A to point R. From R to B the applied flux is increasing in the reverse direction, but the remanent flux does not fall to zero until the applied flux has reached a point between T and B. At point B reverse saturation is reached and from B to Q the remanent flux again shows a flat top. In the absence of any other signal the bias continues to write this pattern.

If the tape is now played back at the same speed the pattern will induce small voltage pulses in the head at bias frequency, but since the pulses are symmetrical about zero, simple integration (to be discussed later) will virtually eliminate them. One thing the bias has established is the maximum wavelength of each pattern on the tape. Larger groupings, which would give rise to low frequency noise if exposed to direct current bias, have been restricted to this wavelength.

It should be noted at this stage that, although the terms amplitude and frequency are used for signal and bias currents, they should not be used when referring to the recorded tape. The pattern on the tape has intensity and wavelength.

SATURATION BIAS

If a low frequency signal, for example audio, is now fed to the head at the same time as bias the pattern on the tape will be modified. Fig. 4 shows the effect, in terms of H and B. The l.f. signal shifts the bias waveform progressively, in this case, to the right. The bias current remains the same in amplitude but now "rides" on the signal. The value of "H", in the right hand direction, increases to well beyond saturation, but the remanence pattern cannot increase with it, as it already corresponds to the saturation value.

However, due to the shift, much more than a half period, in terms of bias, is spent at and beyond saturation, and the pattern on the tape covers a greater distance. On the left hand side, not only is "H" reduced in value, but less than half a period is spent in this region. The pattern on the tape therefore shows a reduction in intensity and in length. The overall bias wavelength has not changed, but the pattern distribution has, and any pulses produced by playback will not be equal and opposite. Integration will show a progressive change in one direction, the value depending upon signal amplitude and the direction upon signal phase.

show the scope for improvement. Assuming a bias frequency of 70kHz and a tape speed of 7in/sec the recorded wavelength will be one tenth of a thou (0.0001in) the pattern representing each flat topped pulse will therefore be one tenth of that. Since the process produces a series of "bar magnets" along the length of the tape, the smallest unit will be the dust particle of the coating, and this will be in the region of one micron, which is about one twenty-fifth of a thou.

The flat top cannot be considered as any more than one particle, and it remains flat with or without a signal. If the dust particles were made smaller, and/or the tape speed increased the recorded length of the flat top could cover many particles and any change of signal level during that passage would modify the top. The slope of the signal would then appear superimposed on the flat top.

PLAYBACK

The surface induction (remanence pattern) on the tape has been shown to be a distorted version of the bias waveform. At very short wavelengths there is a

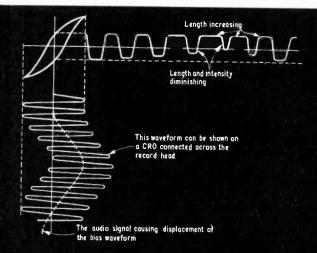


Fig. 4. The effect of applying an audio signal at the same time as the bias signal, producing a saturated magnetisation of tape particles

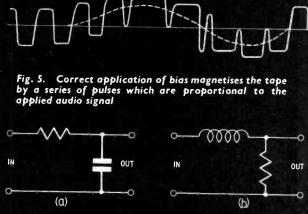


Fig. 6. Filter networks used on playback to eliminate the high frequency bias pulses. A simple CR network is shown in (a) and an LR network using the tape head inductance in (b)

DISADVANTAGES

This system uses saturation bias and has disadvantages. One, the high value of bias tends to demagnetise the pattern recorded, leading to a loss of high frequency response; two, the limiting action described still gives distortion. The optimum value of bias is about half the saturation value, and the remanence pattern on the tape will be lower in intensity. The flat tops will still be present exactly as before, but, when a signal appears, both polarities will be affected, i.e. one half will be increased, the other decreased.

Integration will produce an output whenever the remanence pattern is unbalanced by a signal, but, for any given value of signal, the output will be greater than when using saturation bias because both polarities change in length and intensity.

SCOPE FOR IMPROVEMENT

The system so far described is the typical domestic sound recording, and a few practical details will help to

strong demagnetising effect, and the actual value measurable would be less than expected. The presence of the bias has, however, made the signal effective over the linear part of the characteristic and this is its true function. Fig. 5 shows the effect of a signal on the pattern, the applied signal represented by the dotted line. Each half cycle of signal now appears as a group of pulses.

When the tape is passed over the gap this pattern will produce voltage pulses in the head circuit and the next process can be considered as filtering. Fig. 6a shows a CR network which will act as a low pass filter, suppressing the high frequency pulses, but producing an output proportional to their amplitude/duration. This action is integration, the pulses producing the charging current for the capacitor, the output signal being the voltage developed across it.

Fig. 6b, although less well known, is also an integrating circuit, having the same effect, but the inductance is that of the playback head, usually 0.5 to 1.0 henries.

ERASURE

The advantage of tape is that it can be used many times, and, to achieve this, some means must be used to

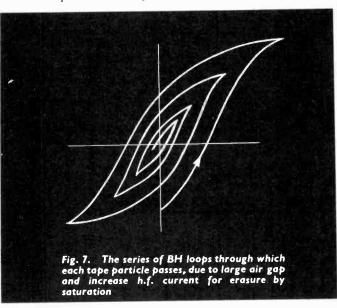
remove the recorded signal.

The most common method is to mount an additional head—the erase head—fairly close to, and just ahead of, the record head. A much larger gap is used and the same bias waveform of greater amplitude, is fed to the winding. The current should be sufficient to take each tape element beyond saturation.

This head is usually tuned to bias frequency since it does not handle signal. Although the gap is large, and the erase current high, it is found that the original pattern on the tape, especially that of strong signals, often recovers after erasure. To avoid this effect a double gap is used, the gaps, in this case, being adjacent

to each other.

Because the gap is larger, and bias current higher, the flux extends well beyond the physical gap, i.e. the fringe is longer. Each tape element spends much more time in the fringe than it would in that of the recording head, and many loops are traversed in that time. The loops fall off in amplitude as each element moves out of



the fringe so that the remanent flux is negligible. Fig. 7 shows the series of loops to which each element is exposed as it leaves the erase gap, the actual number of loops being much greater than a clear sketch could show.

CONSTANT CURRENT RECORDING

When a signal is recorded on tape the pattern must represent the amplitude as well as the frequency. The flux and therefore the head current should be independent of frequency. The recording head is highly inductive and its reactance would reduce the current as the frequency became higher, assuming a constant voltage drive.

A resistor is usually connected between the output of the signal amplifier and the head to reduce this effect, which could lead to a fall-off of 6dB per octave. If the resistor value is high, it will swamp the reactance of the head, so that the head current will be the same for all frequencies of the same original amplitude. This is called a constant current system.

NEWS BRIEFS

B.A.E.C.

The British Amateur Electronics Club recently held its annual exhibition at Penarth, Glamorgan, and as a result of this, some of the exhibits were used in a BBC television programme on BBC Wales; the programme—"Heddiw"—was televised in Welsh but the B.A.E.C. were provided with an English translation of the text that they published in their October Newsletter. The games illustrated on television were all constructed by members of the B.A.E.C.

Vidicon Storage Tube

DEVELOPMENT of the first television camera tube to feature built-in stop-action capabilities was reported recently by RCA.

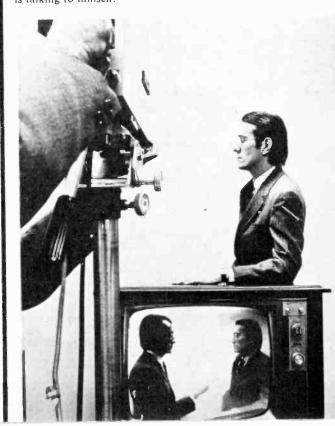
The unit is a small, vidicon-type device that can take—and store for later play-back—single electronic "snapshots" or "stills" of continuous programmes being

transmitted by the tube.

The still picture is electronically stored inside the tube itself thus eliminating the need for an external storage medium such as film or videotape. It can be relayed to a TV monitor immediately, or kept intact within the tube for several days.

The new tube is technically referred to as a silicon storage vidicon and could be used, for example, in a space, satellite to eliminate the bulk and weight of the present video storage devices which relay individual TV pictures to earth by relatively slow communications systems.

The picture below shows the new tube in use. The image on the left of the screen was recorded by the tube a few minutes before, and the one on the right is being transmitted "live", giving the impression that the subject is talking to himself.





By D. S. GIBBS and I. M. SHAW (FERRANTI LTD)

A HIGH quality pre-amplifier must perform several important functions, it must have an input impedance sufficiently high not to load the signal sources, it must provide frequency equalisation where necessary, it must have filters to remove spurious noise or harsh treble from poor recordings and it must have tone controls to accommodate the preferences of the listener, the deficiencies of the loudspeakers and the acoustics of the listening room. Furthermore it should do all this without any significant increase in either distortion or background noise level over a wide range of input signal levels.

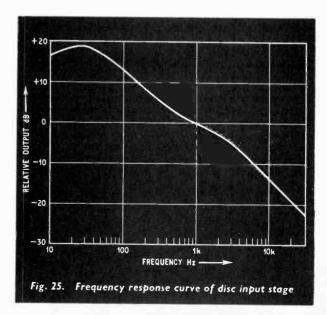
The design evolved meets all these requirements, as the specification published previously shows, and it has several other valuable facilities. A microphone channel is included which can be mixed independently with any other input. This facility is useful to tape recording enthusiasts or can be used as a "home discotheque" for parties etc. Another feature valuable to the tape recorder enthusiast is the recorder output which is taken after the equalisation and filter stages but is unaffected by volume or tone controls. The circuit diagram of the right hand channel of the pre-amplifier is shown in Fig. 24, the left hand channel is identical.

DISC INPUT

The disc input stage comprising T14, T15, T114 and T115, provides frequency equalisation to the RIAA L.P. characteristic with an accuracy of approximately \pm 0.5dB between 30Hz and 15kHz as shown in Fig. 25. The circuit has been designed so that the frequency response curve falls below 20Hz. This attenuates low frequency components in the output of the pickup and makes a separate rumble filter unnecessary. The sensitivity is 3mV in the "Lo" position and about 60mV in the "Hi" position so that either magnetic or ceramic pickup cartridges may be used.

The circuit used is a complementary two stage feedback amplifier, a more detailed circuit of which is shown in Fig. 26. This was chosen in preference to the more popular circuit shown in Fig. 27, because of the greatly improved bias stability and its freedom from the subsonic ringing which the circuit of Fig. 27 tends to produce. This arises through the feed-back provided by R1 becoming positive at certain frequencies due to the phase shifts produced by C1 and C2 (Fig. 27).

The shape of the frequency response curve is defined by the negative feed-back provided by R39, R40 and C19 to 21; the overall gain is defined by R38. The circuit gives an output of 100 mV for an input of



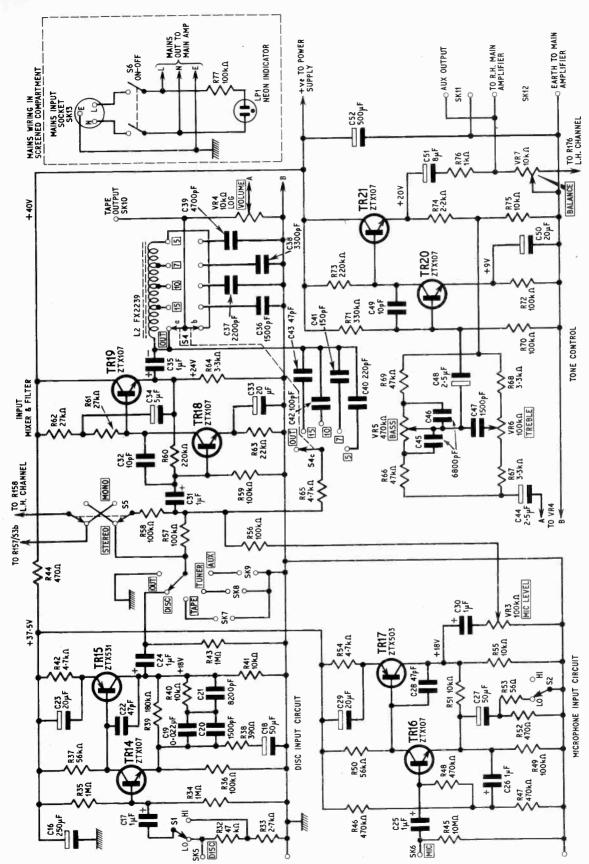
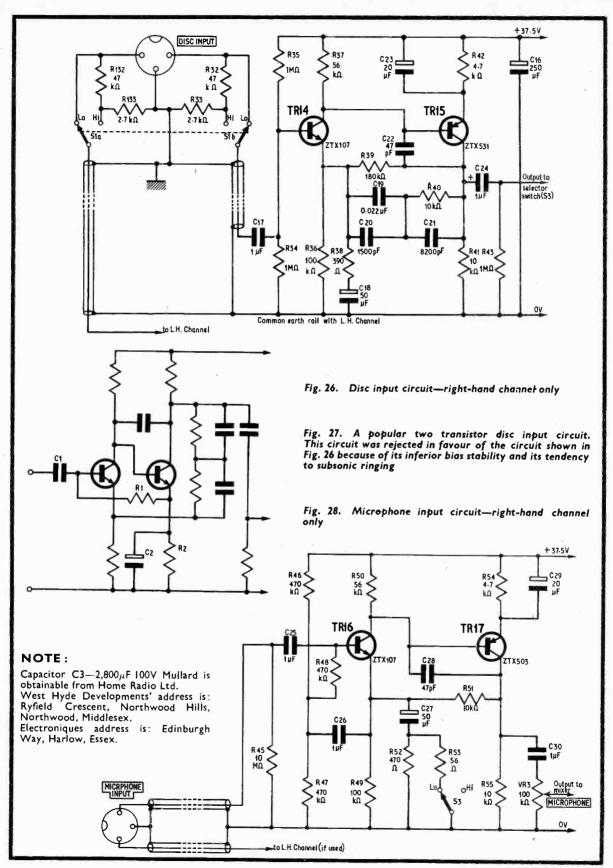


Fig. 24. Circuit diagram of the right-hand channel of the pre-amplifier of the P.E. Gemini. The left-hand channel is identical—component numbers being right-hand channel plus 100, except for the switches and sockets which serve both channels by being double pole types



3mV at 1kHz. Distortion generated by the input stage is below 0.005 per cent with 3mV input at 1kHz and below 0.05 per cent at 20dB overload. Between 100Hz and 10kHz the distortion does not exceed 0.1 per cent at any level up to 30dB overload.

MICROPHONE INPUT

The circuit of the microphone input stage (Fig. 28) is virtually the same as the disc input except that the feed-back network is replaced by a single resistor to give a flat frequency response and the bias resistors at the input are bootstrapped to increase the input impedance. Two input sensitivities have been provided. In the "Lo" position of S2 the input sensitivity is 1mV with an input impedance of 1 megohm which is suitable for most low and medium impedance dynamic and ribbon microphones. With the switch in the "Hi" position the sensitivity is reduced to 10mV and the input impedance rises to 2 megohms. This sensitivity is suitable for high impedance dynamic or crystal microphones.

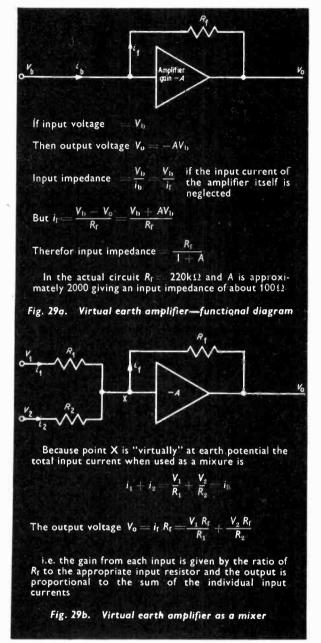
Constructors not requiring the microphone input can simply omit all the components associated with this stage, or if only mono microphone operation is required only one microphone input stage need be built and the output can be fed to both mixers via separate 100 kilohm resistors (R56, R156). Potentiometer VR3 in this case need only be single gang.

The table below summarises the performance of the microphone input stage:

INPUT MIXER

The input mixer works on the virtual earth principle which is explained in Fig. 29. Because the input impedance at the base of the TR18 is low—only about 100 ohms—it can for all practical purposes be considered as a short circuit to earth, hence the name virtual earth amplifier. The total input current when

	Table 1	
	"Lo"	"Hi"
Input Sensitivity	lmV	10mV
Input impedance	1ΜΩ	2ΜΩ
Distortion	less than 0.005% at 1mV input and less than 0.05% at 20dB overload	less than 0.005% at 10mV and less than 0.02% at 20dB over-load
Frequency response	65Hz to 70kHz at -3dB	at -3dB. With 500pF source (crystal microphone) frequency response is -3dB at 60Hz
Signal to noise ratio at rated sensitivity	$-60dB$ with 100Ω source, $-56dB$ with $1K\Omega$ source, unweighted, $20Hz$ to $20kHz$	-70dB with 1k Ω source, -56dB with 50K Ω source, unweighted, 20Hz to 20kHz



used as a mixer is then just simply the sum of the input currents from each signal source, which enables the microphone input to be mixed with any of the other inputs without interaction.

The input sensitivity at the tape, tuner and aux, inputs is 100mV with an input impedance of 50 kilohms, and the output at the emitter of TR19 is 400mV. Capacitor C34 is a bootstrap capacitor which increases both the gain and the available output voltage swing to give minimum distortion. At 1kHz the distortion produced is below 0.005 per cent at rated sensitivity and below 0.02 per cent at 20dB overload, at 10kHz these figures become 0.01 per cent and 0.1 per cent respectively. The circuit will give an output of 10 volts r.m.s. (28dB overload) before clipping occurs. Signal to noise ratio at rated sensitivity is -80dB (unweighted 20Hz to 20kHz).

To be continued

AUDIO TRENDS By F. C. JUDD

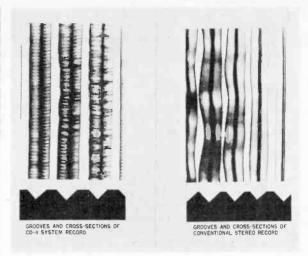
HEN stereophonic sound reproduction made its debut, only a few years ago, the various manufacturers who exhibited at the annual Audio Fair were quick to take advantage of the "moving sound" illusions that could be created with stereo. These consisted mainly of the sounds of express trains and racing cars, etc. which hurtled across the demonstration rooms at breakneck speed, and of phantom ping-pong games in which you followed the movement of the ball without seeing it. After being subjected to this and other subtle forms of audio brain-washing, you went home either cross-eyed or cross-eared but nevertheless convinced that stereo had something even if it was only two of everything.

Now it seems that stereo is out—two channel stereo that is. The "in thing" is four channel stereo alias quadrasonic, alias CD-4 alias surround sound, call it what you like. At least that's one impression gained at the 1970, 16th International Audio and Music Fair held in October at Olympia. This time, no trains and no ping-pong game. This time they sat you down in the middle with the sound all around and intensified the audio brain-washing with real solid state power.

PRACTICAL ELECTRONICS went one better. With their demonstration of the "P.E. Gemini" stereo amplifier, visitors were also treated visually with the "P.E. Aurora", an audio controlled light display, which can be adapted by the constructor for domestic settings. (Details of "Aurora" will be published very soon as a constructional project—Ed.)

FOUR CHANNEL-ON DISC

JVC-Nivico (Victor Company of Japan) recently announced their new four channel disc stereo system and its reproducing equipment known as CD-4. This was demonstrated at Olympia and is claimed to be high fidelity and also compatible with existing two channel reproducing equipment, i.e. the CD-4 four channel discs can be played on conventional two channel equipment and two channel discs can be played on CD-4 equipment.



Comparison of CD-4 disc and conventional stereo record

The four channels are cut in one groove and according to the very limited technical information at present available, the system features both frequency and phase modulation which are combined to carry the signals for the extra channels. Reproduction requires four separate amplifier channels and four loudspeakers. Two loudspeakers are positioned in front of the listener and two at the rear. The photographs show the difference between recording on a two channel stereo disc and the new CD-4 disc. The U.K. distributors are Denham and Morley Limited, Denmore House, 453 Caledonian Road, London, N.7.

FOUR CHANNEL—ON TAPE

The Audiosonic system, which is four channel stereo on tape, was also demonstrated at Olympia by JVC-Nivico. The system employs four separate tape heads and offers various ambionic arrangements for the listener, i.e. two speakers in front and two at the rear or three in front and one at the rear or all four at the front. The makers claim that the system presents truly "solid music" with complete freedom of movement for the listener. As the two recommended twin channel Nivico amplifiers for this system are each rated for 140W music power (total 280W) the system could present some pretty solid sound as well.

Nivico four channel disc player



50 PROJECT ELECTRONIC KIT

Model R.130



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This excellent transistorised battery operated kit will not only provide hours of entertainment when made up but also in its construction. It receives the normal broadcast band 500kHz to 1-6MHz and on short wave 1-5-30MHz in three bands.

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PROFESSIONAL SOLID STATE FOUR BAND

10 WATT BUDGET STEREO ALL TRANSISTOR

from 555kHz to 30MHz
In four bands including
illuminated electronic
bandspread for 160-10
metres. Also incorporated is an internal
speaker, automatic noise
limiter, S8B/AM/CW Switch, AVC Switch, S Meter, Receive and Standby
Switch, external socket for headphone or speaker, bandspread control,
BFO control, on/of/AF gain, band selector, antenna trimmer and RF gain.
The R. 135 will run off of 240V a.c. dry batteries or any 12V d.c. negative
ground source.

(A) *

enclosure. **SPECIFIC.ATION: Output: 10 watts total. 5 watts per channel. **Frequency Range: 35-18,000Hz. Inputs: Phono and Tuner.



Model R.135

20 PROJECT SOLAR ELECTRONIC KIT Mod. R.128

This ultra modern Project Kit is shaped for the space age. Carried inside a transparent domed 4½in capsule the R.128 comes complete with a self-contained solar cell to power any one of 20 projects ranging from a one transistor radio to a morse set complete with key and morse training code. Supplied complete with easy to follow instructions and even the cement to assemble this unique electronic space causule.



PRICE £4.10.0

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Complete with a music book containing 10 casy to play songs the R.129 solid state organ kit covers 2 full octaves. Slotting into a fitted hardwood case the top panel carries the key assemblies and all the components including the loudspeaker. Like all Rec Electronic Kits every item is included down to the last but and boltso that the constructor can start assembly within minutes of opening the package.

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Using a robust Solid State Integrated circuit the R.127 kits will build any one of these projects: (1) Germanium Radio. (2) Test Oscillator. (3) Morse Telegraph Training Set. (4) J.C.J. Transistor Radio. (6) Germanium J Transistor Radio. (6) Record Player Amp. (7) Continuity Tester. (8) AF Signal Tracer. (9) Radio Transmitter. (10) Water Parity Tester.

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2 TRANSISTOR SOLAR RADIO KIT Model R.126

Like all Roc Electronic Kits the R.126 uses reliable no-solder connections to produce a complete 2 transistor radio in under 2 hours. As well as battery operation the kit is supplied complete with a solar cell to provide power from the Sun or any strong light source.

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CRYSTAL RADIO KIT Model R.125

This easy to build Radio is based on the same circuit developed by Marconi for the very first radio transmission but uses a modern ferrite serial for maximum efficiency. A perfect introduction to Radio Theory.

5 WATT STEREO INTEGRATED AMPLIFIER

Model R.123

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AM/FM/MPX STEREO TUNER

(

Model R.134

Perfect Matching Unit to the R.136 Amplifier. The R.134 Stereo Tunor is designed to give years of reliable performance. The Tuning Band covers AM & FM with a separate stereo beacon to indicate when stereo broadcasts are being received

AMPLIFIER Model R.136

Ideal as a second stereo system or for the newcomer to Hi-Fi who wishes to upgrade existing equipment the R.136 sounds every bit as good as it looks. As well as inputs for crystal or ceranic cartridge the R.136 has a Stereo Tuner input which accepts the matching R.134 Stereo AM/FM Tuner.

The Satin Finish front panel carries a stereo diphone socket as well as volume. Balance, tone.

AMPLIPIEK
Another Roc Exclusive offering
top value for money performance
the R.124 is a Sterce Tuner/Amp
with facilities only usually found
in much more expensive units.
Features like separate bass and
treble controls, automatic frequency
control switch and sterce headphone socket give the R.124 a
price to specification ratio second
to none.

AMPLIFIER

PRICE £13.0.0



Model R.124

Mounted on a heavy gauge chassis the fully transistorised R.123 stereo amplifier is completely self-contained even down to ganged volume and separate tone controls. For a simple stereo amplifier of excellent quality all you have to provide is the cabinet and control knobs.

SPECIFICATION:
Output: 5W total. 2-5W per channel.
Input sensitivity: 600mV at 2-2Mohms
AC 240V operation.

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CHASSIS



5 WATT 8 TRACK CARTRIDGE STEREO AMPLIFIER Model R.133

Model R.135

Just slot in one of the many 8 track cartridge tapes available for a continuous programme of your favourite music. A manual programme overtide switch enables you to switch from one track to the next at the push of a button at the same time a numbered indicator lights up to show which track is playing. Beautifully finished in an oiled walnut cabinet the R.133 is mechanically engineered to provide long and reliable service. 8PECIFICATION—Tape speed: 9-5 cm/sec (3\(\frac{1}{2}\)\text{1.0.6.}\)
Wow & flutter: better than 0.3\(\frac{2}{3}\)\text{. Prequency range: 40-12,000Hz.

Cross talk: better than 45dB at 1,000Hz. Output: 5W total. 2-5W per channel. Amplifier outputs: 2000.



s talk: better than 45dB at 1,000 nnel. Amplifier outputs: 200ntV.

PRICE £36.0.0

to none. Housed in a handsome walnut Housed in a handsome walnut cabinet the classical low line styling of the R.124 will grace any home. SPECIFICATION: FM: Frequency Range: 88-108MHz. Usable sensitivity: 20gV. Stereo Separation: 26dB at lkHz, 20dB minimum. Image Rejection: 55dB. AM: Frequency Range: 535-1605kHz. Usable Sensitivity: 300gV. Audio, Section: Total Output Power: 4W. Phone Input: 200mV at IMΩ. Tape Input: 100mV at 100kΩ.

WATT STEREO FM/AM/MPX TUNER

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STEREO HEADPHONES Model R.328

Built up to a standard not down to a price, the R.328 stereo headphones represent a breakthrough in value for

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A valuable addition to any stereo installation they will provide many hours of listening pleasure. SPECIFICATION:

Matching impedance: 8-16 ohms Frequency range: 30-15,000Hz.

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SURROUND STEREO

A cheaper and perhaps equally effective "ambionic" system known as "surround stereo" can be devised using any conventional two channel stereo equipment plus one extra loudspeaker. This system is introduced in a leaflet included with the new EXP-70 LP disc called "This is Stereo".

This record, one of the EXP Technical Series, will by now be available from all record dealers price 39s. It is a combined stereo demonstration and test record containing various recordings that illustrate not only the dynamic and spatial possibilities that stereo offers, but also includes test tracks for adjustment of experimental "surround stereo" systems described in the leaflet and for balance and phasing, etc. of two channel stereo.

SOME NEW AUDIO PRODUCTS

Although over 80 manufacturers were exhibiting at the 1970 Audio and Music Fair, few completely new products were to be seen. Most manufacturers were showing their current ranges of equipment, a good deal of which has already been publicised in the technical press.

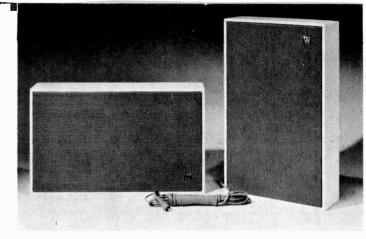
The new Wharfedale Aston wall mounting loud-speaker was one new product and this breaks away from the old hi-fi tradition of dark veneer and rather uninspiring front covering. These speakers are available in glossy white enamelled cabinets and the front is finished with what Wharfedale call their "silver fox" fabric which looks like shiny metal. The Aston speaker system comprises an 8 inch bass unit and 3 inch treble unit and the cabinet measures $19 \times 11\frac{1}{2} \times 4\frac{1}{2}$ inches. Power handling capacity is 18 watts r.m.s. and the input impedance 4 to 8 ohms. These and the new Wharfedale Triton speakers, also released at the same time, are available in matched stereo pairs complete with 15 feet of connecting cables and DIN plugs.

LOUDSPEAKERS FOR THE FUTURE

The loudspeaker has always been the most inefficient link in an audio reproducing system requiring as it does a comparatively large amount of audio power to produce a useful sound output. J. B. Lansing Sound Inc. of America have not only found a way of increasing efficiency but also of producing better sound distribution. Most of their loudspeakers are intended for professional studio use but one known as Aquarius 4 has been designed for domestic use and was introduced at the Audio and Music Fair.

Mounted inside the enclosure is an 8 inch wide range transducer directed upward into a radial horn. This spreads the sound through a 360 degrees horizontal radial defraction slot. The higher frequencies emerge in a 360 degree vertical plane from a 2 inch driver separately loaded by a radial horn mounted at the rear of the enclosure. The result is a much wider distribution of sound than would be obtained with a conventional system.

The Aquarius 4 will handle power up to 30 watts r.m.s. but nothing like this power is necessary for life size sound because of the high efficiency. The same company also manufacture a very compact studio monitor speaker for power capacity up to 75 watts. The U.K. distributors are Feldon Recording Limited, 126 Great Portland Street, London, W.1.



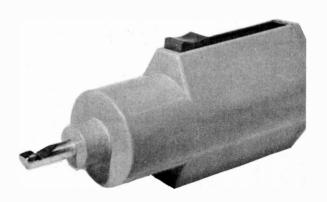
Wharfdale Aston, wall mounting loudspeaker

LESS NOISE PLEASE

The Dolby noise reduction system has, until recently, been available only for studio use. Kellar Electronics Limited have now introduced two Dolby type B noise reduction units for use with any domestic tape recorder having radio or line inputs and line or external amplifier outputs. The KDBI unit is switchable from record to replay and is for use with recorders having a combined record/replay head. This retails at £45. The second unit, known as the KDB2, has two separate record and replay channels for use with machines having separate record and replay heads and retails at £75. Both units are supplied complete with stabilised power supplies and VU meters and are for two channel (stereo) operation.

Kellar Electronics have also released a combined cassette recorder/amplifier known as type DTA50, with a built-in Dolby noise reduction system the retail price of which will be £148 4s 8d. It seems likely that a noise reducing system will become an integrated part of all domestic as well as professional tape recorders. Cassette tape recorders are a very obvious choice to begin with as few have a noise performance of better than -40dB.

Tape head magnetisation is one of the most common causes of high noise level on tape recordings. Tape heads should therefore be frequently demagnetised with a defluxing tool. Together with many other new audio accessories that Ferrograph have just released is a new tape head demagnetiser, their Defluxer type D/2 which retails at £4. The outlay is a very worthwhile



Ferrograph tape head demagnetiser

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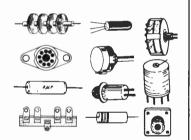
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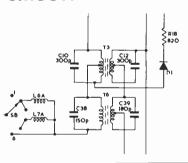
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2N914		2N2906A	6/6		7/-	28005	15/-	ACY44		C152	3/6 BF167	5/- BFW		GET898	4/6	NKT401	17/6	OC70	3/-
2N916		2N2907	8/-	2N3858	5/-	28020	37/6	AD140	8/- B0	C157	4/- BF173	6/6 BPX	25 37/-	MAT 100	6/-	NKT402	18/-	OC71	2/6
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2N930		2N2924 2N2925	3/6 3/6		5/6 6/6	28103 28104	6/6 6/6	AD150 AD161		C159 C160 1	4/- BF178 12/6 BF179	7/- BPY 14/6 BSX		MAT120 MAT121	6/-	NKT404 NKT405	12/6	OC74	6/6
2N987		N2926	0/0	2N3860	6/-	28501		AD162		C167	3/- BF180	7/- BSX				NKT406	15/- 12/6	OC75 OC76	4/6 4/6
2N1131	5/6	Green	2/9	2N3866	30/-	28502	5/6	AF106			2/9 BF181	6/6 BSX	21 7/6	MJ420		NKT451	12/6	0C77	5/6
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2N1304		N3014						AF117		C170	3/8 BF195	5/6 B8X		MJ480	19/6	NKT603F NKT613F		OC81D	4/-
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2N1309		N3134	6/-	2N 3906	7/6	R.C.A.:	EE/0	AF126		C182	4/8 BF224	6/- BSY		MJE340		NKT717 NKT734	8/6 5/6	OC169	4/8
2N1507	5/6 2	2N3135	5/-	2N4058	5/6	40050	13/6	AF127		C184	4/8 BF225	6/- B8Y	II 5/6	MJE520		NKT736	7/-	OC170 OC171	6/-
2N1613		2N3136	5/-		5/-	40250	10/-	AF139		C182L	4/- BF237	6/6 BSY:		MJE621		NKT773	5/-	OC200	6/6
2N1631 2N1632		N 3340 N 3349	19/6 26/-		5/- 4/6	40251 40309	19/6 8/	AF178 AF179		C183L C184L	3/6 BF238 4/- BF257	9/6 BSY:		MPF102 MPF103	* 8/6 7/6	NKT781	6/-	OC201	9/6
2N1637		N 3390			4/6			AF180		C187	5/8 BF22A	9/6 BSY		MPF104	7/8	NKT1033		OC202	9/6
2N1638	7/6 2	2N3391	4/-	2N4244	9/6	40311	9/6	AF181	8/6 BC		4/6 BFX 12	4/6 BSY:	28 3/6	MPF105	7/6	NKT1041		OC203	6/6
2N1639		N3391A	6/-	2N4245		40312		AF186	13/4 BC		5/4 BFX 13	4/6 BSY:		MPS3638	6/6	NKT1043		OC204 OC205	8/6
2N1701 2N1711		2N3392 2N3393	4/-	2N4254 2N4255		40314	9/6 9/6	AF239 AF279	9/6 BC		5/6 BFX29 5/8 BFX43	7/- BSY:		NKT001: NKT124		NKT1051 NKT2032		OC207	12/6
2N1889		N3394	4/-	2N4284		40316		AF280	12/6 BC		5/8 BFX 44	7/6 BSY		NKT125		NKT8011		OCP71	8/6
2N1893		N3402	4/6			40317		AFZII		CY31	5/6 BFX68	13/6 BSY:		NKT126	5/6		15/6	P346A	4/6
2N2147 2N2148		N3403 N3404	4/6			40319		ASY 26		CY 32	7/8 BFX 84	8/- BSY:		NKT128		NKT8011		T1834	12/6 8/-
2N2140 2N2160		N3405	9/-			40320 40323	9/6 8/6	ASY27 ASY28	7/6 BC		4/= BFX85 4/8 BFX86	7/- B8Y- 6/- B8Y		NKT135 NKT137	5/6 6/6	NKT8011	19/6	T1843	2/6
2N2193	9/6 2	N3414	5/6			40324		ASY29	5/6 BC		4/6 BFX87	6/- BSY		NKT210	6/-	AR INOIL	22/6	T1845	3/6
2N2193		N3415	5/6	2N4290		40326	19/6	A8Y36	5/- BC	CY39	8/8 BFX 88	5/- BSY		NKT211	6/-	NKT8021		T1846	3/6
2N2194/ 2N2217		N3416 N3417	7/6 7/6	2N4291 2N4292		40329	7/-	ASY50		CY40	7/6 BFX89	12/6 BSY		NKT212	8/-		18/6	T1847	3/6 3/6
2N2218		N3439	26/-	2N5027		40344		ASY51 ASY53		DY42 DY43	3/- BFY10 3/- BFY11	6/6 BSY 3		NKT213 NKT214	6/- 4/8	NKT8021	18/6	T1849	3/6
2N2219		N3440	19/6	2N5028		40348		ASY54			6/8 BFY17	4/6 BSY		NKT215		NKT8021		T1850	4/6
2N2220		N3570	17/6	2N5029		40360		ASY62	5/- BC	CY 58	4/6 BFY18	6/6 BSY			7/6		18/6	T1851	3/6
2N2221 2N2222		N 3572 N 3605				40361		A8Y63	3/6 BC		4/6 BFY19	6/6 BSYS		NKT217		NKT8021		T1852 T1853	3/6 6/-
2N2287		N3606	5/6	2N5172 2N5174	3/- 10/6	40362 40370		A8Y72 A8Y83			9/8 BFY20 4/_ BFY21	12/6 BSY9 8/6 BSW		NKT219 NKT223	5/6	NKT8021	18/8	T1960	6/-
2N2297	6/- 2	N3607	4/6	2N5175		40406		ASY86	6/6 BC		8/6 BFY24	9/- B8W	70 5/6	NKT224	5/-		18/6	T1861	6/-
2N2368		N3662		2N5176	9/-	40408		A8Z20	7/8 BC	2Y72	3/6 BFY25	5/- D161		NKT225		NKT8021	i	T1P29A	13/6
2N2369 2N2369A		N3663 N3702	7/6 3/6	2N5232A 2N5245		40467A	16/6 14/6	ASZ21 AUY10	8/6 BY		5/6 BFY26 7/6 BFY29	4/- D161 10/- D16P		NKT229	8/-	OC20	18/3	T1P30A	15/-
2N2410		N3703	4/6	2N5246		40468A		BC107	30/- BC 3/- BI		7/6 BFY29	-0/- (D101	0 1/0	NKT237	11-	0020	19/-	T1P31A	16/6
2N2483	5/8 2	N3704	4/6	2N5249	13/6	AC107	6/-	BC108	3/- BI	0121 1	9/_								
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2N2540		N3706 N3707	4/8	2N5265 2N5266	65/- 55/	AC126 AC127		BC113 BC114	5/6 BI 7/6 BI		2/- 9/6	PAYABL	EXCE	PT C.R.T	JP T	DE AF	TER	THATF	REE
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safeguard against the high noise level that can be imparted even to pre-recorded tapes if a tape head has become magnetised.

Another new and useful device suitable for all current Ferrograph tape recorders is a signal operated switching unit which starts the recorder on arrival of wanted signals and stops it again automatically when the signals cease. The device operates from a lowest signal level of 400mV and has an adjustable over-run time of 5 to 25 seconds; retail price £30.

MUSIC AT THE FAIR

A few large musical instrument distributors were showing currently available instruments, such as electronic organs, guitars and amplifiers, but nothing that could be described as new could be found.

One of the highlights in the music field, however, was the Moog Synthesiser which is an electronic music composition and performance system. The system is partly keyboard operated but consists mainly of a large number of voltage controlled oscillators and filters



The Moog Synthesiser — this version would cost between £2,250 and £3,000

which can be interconnected to produce the pitch, waveform and envelope for an infinite range of musical sounds and even of "pseudo" human voices. One of the larger versions of the Moog Synthesiser is shown on this page but such devices are hardly for domestic use since the price range is £2,250 to over £3,000. However, a small synthesiser suitable for schools and small studio use will shortly become available at less than £600. The present U.K. distributors for Moog Synthesisers are Feldon Recording Limited, 126 Great Portland Street, London, W.1.

The designer of the Synthesiser, Dr R. A. Moog, who gave a lecture on and a demonstration of the instrument at Olympia, spoke of the possibility of combining the music synthesiser with a conventional electronic organ resulting in a musical instrument with variable voicing and effects that could be chosen at will by the player. There is already some indication of this idea in modern electronic organs.

AUDIO PRESENTATION PROGRAMME

Visitors to the Fair were also able to hear a series of lectures (many with demonstrations) by well-known

experts in sound reproduction and music. Among these were Dr R. A. Moog mentioned above, P. J. Baxandall, inventor of the almost universally used Baxandall tone control network, Desmond Briscoe from the BBC Radiophonics Workshop and Sir Arthur Bliss (Master of the Queen's Music). Many well-known audio journalists and designers also gave talks on high fidelity sound reproduction. Other special features included a recording studio where visitors could make a record, a search for a singer contest, a music tuition bookshop and a comprehensive display of musical instruments of all kinds.

A number of films of interest to audio enthusiasts such as "The Timeless Track" and "The Magic Tape" by B.A.S.F. and "Sound on Tape" by Philips Limited, were also shown during the week of the fair. All this, together with the displays and demonstrations of products by over 100 exhibitors made the 1970 Audio and Music Fair a successful and interesting venture for both the exhibitors and the visitors. However, the exhibiting manufacturers of audio equipment still had the problem of accommodating a worthwhile number of people during demonstrations; the demonstration cubicles were just not large enough.

NEW RELEASES

Rola Celestion Limited—a new loudspeaker—the Ditton 120. The 120 was displayed at the Audio Fair; it consists of three units, an HF1300 treble speaker, a long throw mid range and bass speaker and an audio bass resonator (A.B.R.) all housed in a case measuring $17 \times 7\frac{3}{4} \times 8\frac{3}{4}$ inches. Frequency response is 30Hz to 15,000Hz, power handling is 20W DIN, impedance is 4 to 8Ω ; the speakers are available in matched pairs and cost £48 per pair.

K.E.F. Electronics Limited—a new range of loudspeakers known as the Cresta, Chorale, Cadenza and Concerto and each has a three speaker system. The Cadenza was released at the Audio and Music Fair and is a completely new system with a passive bass radiator. Frequency range 30 to 30,000Hz. Power handling capacity 25 watts at 8 ohms.

Uher-Bosch Limited—a new tape recorder, the Uher 724, which is a quarter-track stereo machine for 7½ and 3½ i.p.s. Specification is to DIN standards with an output power of 2 watts per channel. The recorder operates in the vertical or horizontal position. Retail price is £103 10s 0d plus £25 11s 5d p.t. Uher have also introduced a new portable cassette recorder for C60, C90 or C120 tapes and which features a reversible drive system to obviate having to turn the tape cassette over. Price not available at time of writing.

B.A.T.R. CONTEST

Prizes were presented at the Audio and Music Fair to winners of the 1970 British Amateur Tape Recording contest. The "Tape of the Year" winner was Mr K. McKenzie for his documentary tape called "Sunderland Hospital Broadcasts". For this he was awarded the Emitape Challenge Cup and also the Philips Shield for the best recording in the documentary class.

The 1971 and 14th British Amateur Tape Recording Contest has already been announced with a closing date of 30th June, 1971. Entry forms are available from the Secretary, B.A.T.R. Contest, 37 Fairlawnes, Maldon Road, Wallington, Surrey. As before there are six categories to choose from including one for schools.

By R. W. Coles Part 2

The alarm board "B" carries the logic and oscillator circuitry required by the alarm system, and the 5 volt regulator, which provides the highly stable supply required by all the i.c.s used in the clock. The board is physically smaller than the main clock board "A", being a "Dualine" DL109/22, which has positions for nine 14/16 pin DIL packages, and is equipped with a single-sided 22-way edge connector.

Six i.c.s are used in all, but two of these are CA3046 transistor arrays which are 14-pin packages containing five individual transistors. These are not members of the TTL family used to perform all the clock and alarm

logic.

ALARM OPERATION

The time at which the alarm is required to sound is entered by means of a bank of thumbwheel switches S2a and S2b mounted adjacent to the display on the front of the clock.

These switches are so constructed that they give a four-bit binary coded decimal output pattern for each number selected. This b.c.d. word is continuously compared with the b.c.d. output from the clock counters in a digital comparator (see Fig. 7). When the switch setting and the display are the same, the output of the comparator falls to a low level and is used to set a simple "latch" bistable in IC18.

The output of the latch enables a gate G5c which allows the alarm tone through to a 40 ohm moving-coil speaker, and thus the alarm is sounded. The alarm will

continue indefinitely unless reset by means of the miniature toggle switch S3, also mounted on the front of the clock.

The alarm setting accuracy could be extended to any degree by simply adding more switches and increasing the comparator size. A setting to within 0·1 second is quite possible if required, but with the prototype the system was intended for domestic use, and time setting in ten minute increments was considered adequate.

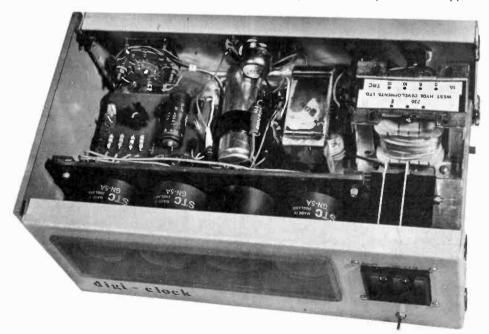
It was also considered unnecessary to include the hours 10.0 to 12.50 in the comparison, thus saving an expensive switch bank which would have been necessary in the "tens-of-hours" position. If this expense is unimportant, or if the aesthetic consequences of using a toggle switch in this role (for this is all that is required in this position) can be tolerated, these times are easily added because the comparator as described has allowance for this extra switch.

There is therefore a considerable amount of flexibility in the alarm circuitry which allows individual constructors to tailor the system to their own requirements.

ALARM CIRCUIT

The circuit of the alarm system is given in Fig. 7, and it can be seen that it is not very complicated at all. Of course, the alarm circuit can be omitted altogether without affecting the remainder of the clock circuit.

The digital comparator is formed from two SN7483 MSI packages, each containing four complete full-adders, with internally connected ripple through carry.



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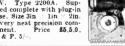
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Each package has four A inputs, four B inputs, four "sum" outputs and a "carry-in" to the first adder and

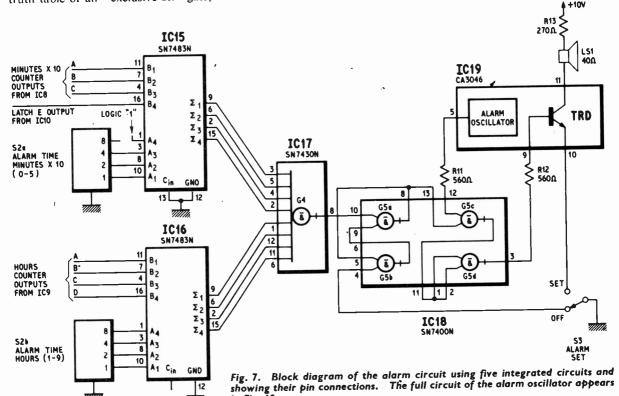
"carry-out" of the last.

These circuits are mainly intended for high-speed multiple bit parallel addition in computer arithmetic units, but binary adders are versatile devices, capable of performing several logic operations. The availability of four such devices in one package opens all manner of possibilities to the logic designer.

To understand exactly how these circuits can be used as comparators, consider Fig. 8. Fig. 8a shows the truth table of an "exclusive-or" gate, which is some-

should be equal to the switch outputs before the completed comparator registers an equality. To achieve this, the eight sum outputs from the adders (each of which registers equality of one digit) are fed to an SN7430 8-input NAND gate G4 in IC17, the output of which will only fall to a logic 0 when all the individual comparisons are valid.

The output of this gate will remain low for ten minutes, until a further increment in the "tens-of-minutes" count invalidates the comparison. A period of this length would probably be sufficient for alerting the sleeper, but to make absolutely sure that the alarm



in Fig. 10

times called a "non-equivalence" gate because its output is a logic 1 only when its inputs are exactly opposite.

To use such a gate as an equivalence circuit which gives a logic I output when its inputs are identical, it is only necessary to invert one of those inputs by means of an inverter gate, or if complementary inputs are available anyway, to compare the true form of one input with the negated form of the other. It follows then that this type of gate could be used to build the complete alarm comparator we require.

As Fig. 8b shows, the truth table of a full-adder is identical to that of the "exclusive-or" gate, if we ignore the carry output and keep the carry input at 0. In fact, using the SN7483 as four "exclusive-or" gates suits our comparator design admirably. The inverted form of one of the words to be compared is already available, because the thumbwheel switches give outputs in this form. The fact that we do have a carry input available can be put to good use in correcting the lagging output of the hours counter.

EIGHT DIGIT EQUALITY

So far this article has only considered the comparison of single digits, but all eight digits from the counters

cannot be ignored, the output of IC17 sets a latch formed from two cross-coupled 2-input NAND gates G5a and G5b in IC18, which ensures that the alarm tone will continue until reset, even if this process takes longer than ten minutes.

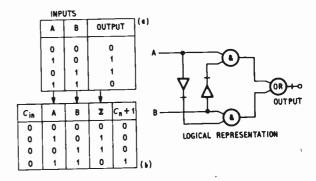


Fig. 8a. Truth table for "exclusive-OR" function

Fig. 8b. Truth table for full adder when Cin = 0

ALARM TIME SWITCHES

The switches used to set the alarm time were chosen from the Birch-Stolec "standard" range, and both banks are coded S.B.10.N1248. These switches are well suited to this application because they give an output on four lines which conforms to the binary code, a separate pattern being produced for each of the ten positions of the numbered wheel.

Any number of switch banks can be mounted together if the constructor wishes to expand the alarm system, the final assembly being finished off by means of a pair of end cheeks which give a neat appearance

and provide the mounting holes.

The principle on which these switches are based is that of a four-track printed, gold plated stator, traversed by four phosphor bronze rotor contacts. The life of this assembly is at least 100,000 operational cycles when switching 24 watts, so in this application the life will be much extended due to the very small current being carried.

To fabricate a similar system using wafer switches, a four pole ten-way switch would be required in each position, with a considerable amount of wiring necessary

to programme the required code.

The wafer switch system does provide a useful way of explaining the action however, and a diagram with output logic table conforming to the thumbwheel switch coding is given in Fig. 9. Note that the output is in the complement form (assuming positive logic). A logic 0 is represented by a ground connection, and a logic 1 by an open circuit.

DIGITAL COMPARISON

As discussed earlier, this inverted code is just what is required for connection to the comparator. Of course, numbers one to nine only are required as inputs to the hours comparator, and numbers zero to five for the tens-of-minutes circuit. Stops could be fitted to the switches if desired, although this is not really necessary; blanking the undesired numbers with paint would be a simple alternative.

Only six positions are required for the tens-ofminutes switch. It follows, therefore, that the full output of four digits is not needed. In fact, the "8" output from this switch can be ignored in the comparison, leaving a spare comparator section in IC15, which is put to good use as an "inhibit" while the "tens-of-hours latch" E output is high. Remember that the alarm system is not operative during this three-hour period as

it stands.

This inhibiting logic is simply achieved by comparing the E line with a permanent "1", thus only permitting a 1 output from this section when the E line is at 0.

As an alternative, it would be possible to ignore this spare comparator section, and feed the E output from IC10 (Fig. 2) to the 8-input gate in IC17 which, when using this method, would have an input available. This system was not used in the prototype simply to retain the design simplification of feeding only true outputs from the main clock board.

HOURS COUNT CORRECTION

So far so good, but up to now we have ignored the point made last month that the hours counter outputs do not conform directly to the binary code, 0000 representing decimal 1 instead of 0, and so on up to 1000 representing 9 instead of 8.

The thumbwheel switches, being "off-the-shelf" items, do not allow for this idiosyncrasy. The comparison as previously described would reveal that, when

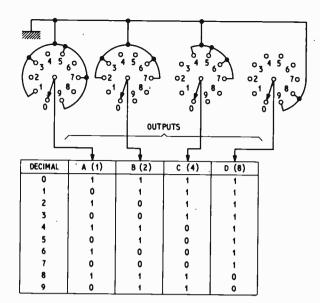


Fig. 9. Wiring of a 4-pole 10-way switch to give a four-bit binary coded decimal output in complement form. The truth table shows positive logic

the switch was set to, say, four o'clock, the alarm would sound at five o'clock, and so on, a most unhappy state of affairs.

It is evident, however, that the code output from the hours counter could be corrected by simply adding binary 0001 to it, increasing its binary value by one. This simple conversion could be carried out using a four-bit parallel adder such as the SN7483N. (A fourbit adder is necessary because carries must be allowed for over all four bits.)

A separate SN7483N could be employed to carry out this correction, but as these circuits are used as comparators anyway, this is not necessary. A I can be added in by feeding a carry in on the "carry" input to the least significant adder. Admittedly this is all a little difficult to grasp, but it does work, and the best way of proving it is to set an example down as a sum, thus:

Hours display: 7, Switch setting: 7. Binary 7: 0111 Therefore switch output (inverted): 1000) Hours counter output: 0110) 1 (plus 1 carry) 1111 equal (alarm sounds)

RISK OF TRANSIENT OPERATION

During the "paper" design of this part of the clock it occurred to the author that, due to the relatively lengthy propagation delay of the SN7483N, and the non-synchronous inputs from the MSI ripple counters. it might be possible for all the adder outputs to be at 1 when the comparison should not show equal, thus giving a transient output from 1C17.

As any such transient "low" output longer than a few nanoseconds is likely to set the latch and sound the alarm (at the wrong time), this state of affairs must obviously be avoided. A good deal of thought was TRANSISTOR RADIOS TO BUILD YOURSELF

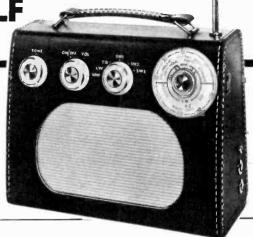
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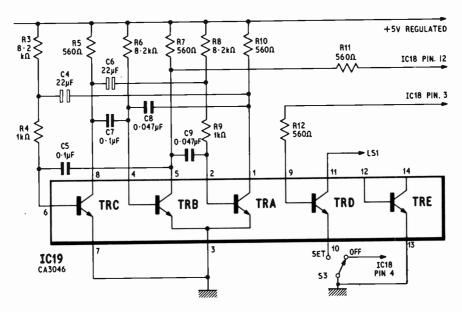


Fig. 10. Alarm oscillator circuit using an integrated package of five transistors. TRE is unused

put in to discover whether this could in fact occur. It was decided that the only way to find out was to "try it and see"!

In the event, it was found impossible to generate this condition in the completed clock, and so all was well, but the point has been made because, with the wrong combination of adders and counters, it was thought to be a possible occurrence.

If any reader should be unlucky enough to experience this phenomenon, the solution is simply to gate the comparison with a clock waveform from one of the divider stages not used in the comparison. In this way the output from IC17 could be strobed only after the hours and tens-of-minutes counters have changed and the SN7483N "carries" have propagated. Obviously some circuit modifications would be necessary to achieve this.

ALARM OSCILLATOR

The alarm oscillator circuit is shown in Fig. 10, which may attract the interest of those who are not otherwise concerned with the clock, because of its obvious applications in mahy other designs.

The circuit was described by A. B. Blackwell-Jones in a letter to Wireless World (August 1967). He called it a "Trivibrator" or "Donkey Simulator", a very apt pair of names indeed. The circuit is based on the usual two transistor astable multivibrator arrangement, to which an extra transistor has been added, so that three separate cross-coupled timing networks result instead of the usual one, as shown in Fig. 11.

Each of the timing networks has a different natural frequency, two being audio tones of about one and two kilohertz respectively, and the other being a low frequency timing circuit giving a 5Hz gating effect. The output is taken from the centre transistor of the two a.f. arms, and consists of alternate bursts of the two tones, each lasting for about 0.2 second, controlled by the l.f. arm. When fed to the speaker this tone gives a very pleasing "space age" warble effect.

If the reader is familiar with the basic multivibrator circuit action, there should be no problem at all in

understanding this extension of the idea. It will help to refer to Fig. 11.

CIRCUIT OPERATION

Either transistor TRA or TRC will be saturated at any time. This saturation will last for about 0·1 second, as controlled by the 5Hz timing network coupling them. The saturated transistor is unable to react via its a.f. timing network with TRB during this time, so this tone is not produced. Meanwhile, however, the other transistor (of the pair TRA or TRC) is not saturated by the l.f. arm and is free to interact via its a.f. timing network with TRB, giving an a.f. output of the appropriate frequency. When the 0·1 second period has expired the l.f. arm changes over, saturating the other (TRA or TRC) transistor, and allowing the other a.f. tone to be produced, and so on.

The only unfamiliar parts of the circuitry are the two resistors, R4 and R9, which are necessary in the trivibrator to prevent the very low impedance of the l.f. coupling capacitors shunting the a.f. arms. In effect, the capacitors C4 and C6 coupling this section are taken to taps on the CR timing resistance chain.

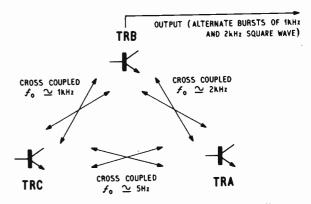


Fig. 11. Principle of operation of the alarm oscillator

The output of the oscillator is taken from TRB via a resistor R11 to gate G5c (Fig. 7), where it is held until allowed through by a logic 1 input from the alarm latch G5a. Note that the alarm oscillator runs continuously, whether the latch is set or not.

When G5c is not enabled, its output will be at a permanent 1, or positive, level. If this output is fed to the simple saturating switch used to drive the speaker directly, although the alarm would not sound, this transistor would be hard on, giving a high power drain and possibly damaging the speaker. To avoid this, G5d is interposed between G5c and the switch TRD to act as an inverter, ensuring that when the alarm circuit is inactive, the output transistor will be off, and no current will flow through the speaker.

The three transistors forming the oscillator, and the transistor used to drive the speaker, are all contained in an integrated package IC19. The other transistor in this array is not used, but it is vital that its emitter (pin 13) should be grounded, as it is connected to the substrate of the device. An open circuit on this pin would jeopardise the isolation between the separate

transistors.

The positive supply for the a.f. output stage is taken from the unregulated 10V line, not to gain any power advantage, but to isolate the 5V line powering the logic circuitry from any ripple produced.

The emitter of the output stage is grounded through the combined "alarm-off/reset" switch S3. When this switch is in the "off" position the alarm cannot sound under any conditions, and a 0 is fed to G5b to reset the latch ready for the next alarm setting.

Note that during the ten-minute period that the alarm output from IC17 is active, it is not possible to reset the latch by simply flicking the switch up and down.

This concludes the description of the alarm circuitry; the 5V regulator, which is wired up on the same board, follows next. It is recommended that the 5V regulator be wired up before the alarm circuit as will be realised later in this article.

FIVE VOLT REGULATOR

The 5V regulator uses a total of seven transistors, five being contained in another CA3046 (IC20), one being a discrete 2N706, and the last being a plastic encapsulated power transistor, type MJE521, which is mounted off the board using the chassis as a heat sink.

The regulator features fold-back current limiting, excellent temperature stability, and an output impedance of less than 0·10hm from d.c. to 100kHz together with line regulation of better than 10mV per volt.

Inputs to the circuit are 10V d.c. (nominal) and 200V d.c. (nominal), provided by conventional power supplies which are described next month. The regu-

lator circuit is shown in Fig. 12.

The 200V line is reduced to provide a 12V biasing supply set by a discrete Zener diode D3, which is bypassed by C1 to suppress any high voltage transient surge when switching on. The 12V is used to power the temperature compensated 7.5V reference supply and the differential amplifier. Using a separate bias supply rather than the unregulated 10V to power these sections gives a great improvement in line and load regulation, and increases ripple rejection.

In the alarm oscillator circuit the advantages conferred by employing a transistor array were those of compactness and convenience, but in this circuit the benefits of the monolithic construction provide a specification which could not be equalled simply by replacing the array with discrete devices. Because

each transistor is mounted in close thermal contact with its neighbours, tight temperature tracking is assured, enabling the construction of an accurately compensated reference and a differential amplifier with a negligible input offset voltage temperature coefficient.

As each transistor was made at the same time and experienced the same diffusion process, parameter matching is also assured. The transistors' $V_{\rm be}$ are guaranteed to be within 5mV of each other; all these factors can be put to good use in this type of circuit.

REFERENCE SUPPLY

The reference supply is provided by TRD and TRE, which are not used as transistors in this application, but as a Zener diode and forward diode respectively. The use of a base emitter junction as a Zener is not new, and the transistors in the CA3046 provide a breakdown voltage of about seven volts when used in this way, with a temperature coefficient of plus 2mV/degree C.

A temperature induced variation of this order is not serious in itself, but if it occurs at the same time as other variations, it could take the 5V supply outside the necessary design limits of plus or minus 250mV.

To eliminate this drift, a forward biased diode, formed from another base emitter junction (TRE), is connected in series with TRD. A junction biased in this way exhibits a forward voltage of around 600mV, but it has a negative temperature coefficient of 2mV/degree C, which cancels the plus 2mV/degree C of the TRD "Zener" junction.

A variable potential divider is used to tap off an accurate 5V from the reference line, and is used as an input to the non-inverting input of the differential amplifier (TRA and TRB). A capacitor C11 is used to bypass any noise which may be present on this

supply

Because a 5V reference is used, the differential amplifier can be used in the voltage-follower mode, with the final 5V output being fed back directly to the inverting input without the usual potential divider chain being interposed. This connection gives a maximum value of loop gain for good regulation, and also increases the frequency response of the amplifier, giving good transient response.

DIFFERENTIAL AMPLIFIER

A differential amplifier was used because of its high voltage gain and absence of temperature problems

which affect single-ended designs.

The output of the "error" amplifier is taken from the collector of TRA and used to drive the base of TR1 which, with the series pass power transistor TR2, forms a compound emitter follower with a very high current gain, and consequently low output impedance. TR2 has to pass all the current required by the load, which could be as high as IA, and is therefore a power device bolted to a heat-sink.

FOLDBACK CURRENT LIMITING

The foldback current limiting circuit is provided by TRC with R21, R22, and R23. The following para-

graphs describe its operation using Fig. 12.

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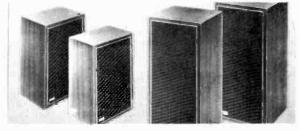
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2N706A	2/6	AF126 AF127	3/6	BFY51	4/-	NKT277	4/-	OC84	5/-
2N708 2N709	3/- 12/6	AF127 AF139	3/6 6/-	BFY52	4/6 3/6	NK T278	5/- 6/-	OC114 OC122	7/6
2N711	7/6	AF178	9/6	BFY64	8/6	NKT277 NKT278 NKT301 NKT304 NKT403	7/-	OC122	10/- 10/- 5/-
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2N1131	6/-	AF181 AF186	8/6	BSX 50	18/6	NKT713	5/-	OC141 OC169	12/6
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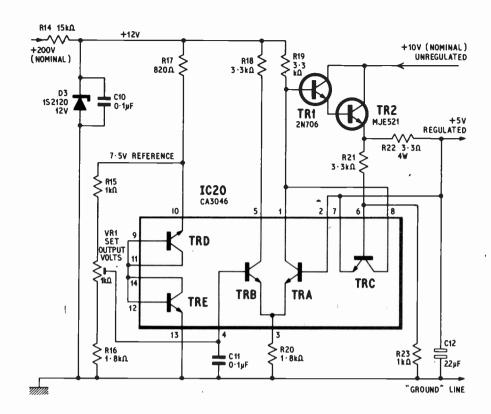


Fig. 12a. Circuit diagram of the 5V regulator circuit including foldback current limiting

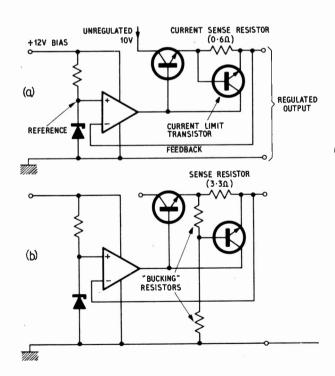
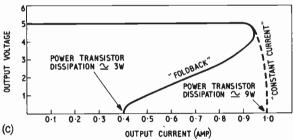


Fig. 12b. Skeleton circuit of a 5V regulator with constant current and foldback current limiting



is easily provided by adding a current limiting circuit, of which there are numerous designs available, the simplest being the one transistor "constant current" arrangement.

A skeleton circuit of the regulator with this kind of limiting is shown in Fig. 12b, along with its limiting characteristic. The operation is quite straightforward; a current sense resistor in series with the 5V supply is used to develop a voltage proportional to the current drawn.

When this voltage approaches the $V_{\rm be}$ of the current limit transistor, it turns it on, diverting some of the base drive to the series pass transistor, and causing the output voltage to fall to maintain the preset current. If the load on the supply is reduced to zero ohms (i.e. a short circuit) the output voltage falls almost to zero allowing only a current preset by the sense resistor value to flow.

The short-circuit current must be set to greater than the permissible load current. For this particular regulator this would be up to 1A, giving a very high dissipation in the power transistor under short circuit conditions, because most of the unregulated voltage appears across this device.

HANDLING THE HEAVY LOAD

This problem could be overcome by using a heat sink large enough to handle the extra dissipation under fault conditions, but there is a much simpler method that is used in this regulator and requires only the addition of two resistors.

In Fig. 12b these extra resistors have been added, and are used as a potential divider to "buck-out" the voltage developed across the sense resistor, preventing it from turning on the transistor at the previous current level.

With the resistor values used in this regulator, the limit transistor will not be able to turn on until the

COMPONENTS...

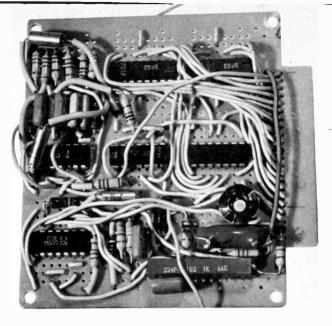
ALARM & 5V REGULATOR BOARD (a) Alarm Oscillator (b) 5V Regulator Resistors R3 8·2kΩ R10 560Ω R14 15kΩ R21 3·3kΩ R22 3·3Ω 4W R11 560Ω RI5 IkΩ R4 IkΩ R12 560Ω RI6 I·8kΩ R23 IkΩ R5 560Ω R17 820Ω R6 8·2kΩ R13 270Ω R7 560Ω R18 3-3kΩ R19 3-3kΩ R8 8·2kΩ R9 IkΩ R20 1-8kΩ Potentiometer VRI Ik Ω linear helical (Painton Bourns type 224P-1-102) Capacitors CIO 0· IμF met. foil 250V 22µF tantalum 15V 0·IμF met. foil 250V CH 0·1μF met. foil 250V 22μF tantalum 15V C12 22µF tantalum I5V met. foil 0.047µF 0-047μF) 250V (Mullard C281) **C8** C9 Diode D3 IS2120. 12V 400mW Zener **Transistors** TRI 2N706 TRS MJE521 (with mica washer) Integrated Circuits ICI5 SN7483N (BP83) IC16 SN7483N (BP83) IC17 SN7430N (BP30) IC18 SN7400N (BP00) IC19 CA3046 (R.C.A.) IC20 CA3046 (R.C.A.) Further notes on purchasing i.c.s given in the article "Making the Most of Logic ICs" last month Miscellaneous S2 2-bank printed thumbwheel switch 10-way (see text) (Birch-Stolec or Radiospares)

S3 Miniature toggle switch, single-pole, changeover

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current drawn approaches 1A, instead of the 250mA level set with constant current limiting and the same sense resistor.

When the transistor does turn on, however, the output voltage will drop, removing the effect of the 'bucking' voltage and reducing the output current. When the output load becomes a short on the output, the final voltage will have dropped almost to zero, and the output current fallen to less than half the value at which limiting began.

This effect is termed "fold-back" limiting, for obvious reasons, and cuts the dissipation in the power transistor, under short circuit conditions, to about a third of that obtained with "constant current" limiting, operating at the same maximum current—a very worthwhile

return for a couple of resistors.

In the prototype the limiter began to operate at about 1A, and with a shorted output the current dropped to 400mA, although resistor tolerances will affect these figures with later versions. The current limiting transistor TRC, is contained within the CA3046, which gives the added advantage that the operating and shortcircuit current will decrease as the chip warms up, due to the negative temperature coefficient of the device's $V_{\rm be}$, giving a degree of thermal feedback.

OUTPUT VOLTAGE CONTROL

The output voltage of the regulator is not affected by the limiting circuitry under normal conditions, because the feedback to the error sensing amplifier is taken from after the current sense resistor R22, the voltage drop across this resistor being automatically allowed for by the amplifier.

The capacitor C12, which should either be a $22\mu F$ tantalum type or an aluminium electrolytic in parallel with a $0.1\mu F$ paper capacitor, reduces transient spikes on the output and also ensures that the regulator

remains stable at all times.

Using a miniature helical preset potentiometer for VR1 allows accurate setting of the output voltage. This type of component also features excellent temperature and long term stability. There is, however, no overriding reason why a single turn rotary control should not be substituted as an economy if a sacrifice of these qualities can be tolerated.

The construction of the regulator should precede that of the alarm circuitry; details of these will be given next month before describing the main power supply.

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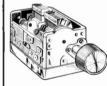
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						e note below)
C	1/20W	5 %	82 Ω -220k Ω	E12	18d	16d	15d
C	₹W.	5%	4·7 Ω -330k Ω	E24	2 5d	2d	1.75d
C	₹W.	10%	4·7 Ω -10M Ω	E12	2-5d	2d	1·75d
C	ł W	5%	4·7 Ω -10M Ω	E24	34	2·5d	2·25d
C	īW	10%	4·7 Ω -10M Ω	E12	6d	54	4-5d
MO	1 W	2%	10 Ω -1M Ω	E24	9d	8d	7d
WW	iw	$10\% \pm 1/20 \Omega$	0.22 Ω -3.9 Ω	E12	1/6d all c	uantities .	
WW.	3 W	5%	12 Ω – 10k Ω	E12	1/6d all c	nuantities	
WW	7 W	5%	12 Ω -10k Ω	E12	1/9d all c	uantities	
CODES	C and an	Alex blob stabilities loss		ovida Floatrosi			WW - wire

CODES: C = carbon film high stability low noise. MO = metal oxide Electrosii TR5 uitra iow noise. No = wise wound Plessey.

VALUES: El2 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, 43, 51, 62, 75, 91 and their decades.

Prices are in pence each for same obmic raine and power rating, NOT mixed values. (Ignore fractions of 1d on total resistor 250 20%: 0.01, 0.022, 0.033, 0.047, 8d ea.; 0.068, 0.1,

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	Log/antilog	10K, 47K, 1M Ω only	8/6
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BATH GOSSIP

"Daddy", chorused my two girls the other day, "Mummy says you're trying to talk through your bathwater. Is that true?" Well how can you best answer this sort of question when you have the bath half-full of tepid water and a whole maze of electronics and tangled wires connected to electrodes dipping over the side!

"You look as if you're going to 'lectracoot yourself", they continued, quite unconcerned about the unanswered question, "Can we help?" Since I had no immediate desire to meet an early demise, I thought it prudent to explain just what was

going on.

Actually, I had for a long while been toying with the possibility of using one of Marconi's earlier ideas for transmitting through water. His technique, which he later abandoned, was to utilise a pair of conductors on one side of a river connected to a carbon microphone and battery, and another couple of electrodes on the opposite bank hooked up to a pair of headphones. This arrangement essentially amounts to a bridge circuit, which, provided the provided impedances between the various electrodes are maintained as in Fig. 1, some fair results can be obtained, but only fair. Put a little two stage amplifier in at the transmitter and another in at the receiver

and you have something that will "go-out" quite a way. Try it and see.

Two hours and four ounces of liquorice comfits later, and my bathtub project, still not off-the-ground, saw me leaving this silent epitaph of despair for the more successful reaches of the Basingstoke Canal.

WALKING MACHINES

Who, you might well ask, wants a machine that walks! Particularly there are already such since perfectly satisfactory devices as the wheel, and even caterpillar tracks for more difficult terrain. Well, it seems that at least two groups of people are extremely interested in having a viable mechanised walker at the earliest opportunity. One, the Army Tank Automotive Command in Michigan, have already got an experimental "clomper" to play with, and I understand that such a machine is likely to be brought into use where wheels or tracks leave off. This pretty sizeable piece of "ironmongery" is apparently capable of climbing over objects in its path and walking through swamps.

The Army machine weighs 1,400 odd kilogrammes and each of the powerful legs is operated by an hydraulic "muscle" driven by highpressure oil. Incorporated within the legs are tactile sensors which permit a degree of force-feedback control, so enabling any forces encountered by the machine to be sensed and dealt

with accordingly!

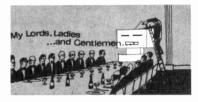
Likely to have a more valid interest in walking devices are the gentlemen of the medical profession, who for years have sought the perfect walking prosthesis with precious little success. However, there seems no reason why hope should be completely abandoned because Prof. A. A. Frank of the University of Wisconsin (who has already completed a very successful four-legged creation) is currently engaged in work on a machine having two legs. This set-up, he forecasts, is anticipated to take the place of the conventional wheelchair and, in time, would give

amputees more freedom than they currently enjoy.

Coordination of the legs in Frank's device is envisaged to be under the direction of a miniaturised digital This, basically, will computer. consist of a random access core memory which can be addressed by read/write associated registers. Initially, the handicapped person would need to make the rather tedious walking movements via the artificial legs himself, but once a satisfactory gait had been established the data could be written into the memory for subsequent use. From then on, whenever walking was required, the memory would be read in a continuous fashion, rather like a tape-recorder using a loop of tape, so producing cyclic leg action; left, right, left, right, and so on until they were required to stop or maybe perform some other function.

A working prosthesis is expected to be in existence round about 1972; it would be nice to think that such worthwhile devices will be capable of being mass-produced fairly soon. Without a doubt the demand will be

high.

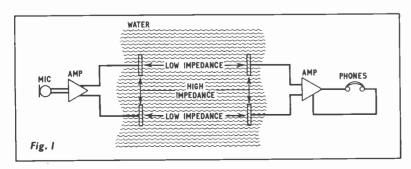


CHATTY COMPUTER

For 'umpteen years now, workers in the field of speech research have hankered for a computer that could produce good synthetic speech without the prior need for human translation of input data into "machine-language" before it could "understand". They need hanker no more, for scientists at the Bell Laboratories in New Jersey have recently taught a computer to convert printed English directly into synthetic speech.

During the experiments the computer was programmed with a basic vocabulary of words, together with their categories and definitions plus such related data as the pitch, stress, and timing as used in average normal conversation. In order to achieve the most natural sound, the computer was additionally given mathematical approximations to the changes in shape which occur in the human vocal tract when making speech sounds.

In use, information for conversion to synthetic speech is fed to the computer via a conventional teletype machine. Since the programme incorporates stress data, one must trust that its digestion is not upset too greatly by such word-group similarities as "a noisy noise", and "a noise annoys", or, dare I say it, "a noisy noise annoys". "Burp!"



Selection from our **po**stbag

Correspondents wishing to have a reply must enclose a stamped addressed envelope. We regret we are unable to guarantee a reply on matters not relating to articles published in the magazine. Technical queries cannot be dealt with on the telephone.

"P.E. Gemini"

Sir-I unfortunately do not know whether Mr D. S. Gibbs or Mr I. M. Shaw was responsible for writing the introductory article for "P.E. Gemini" Stereo Amplifier, but this preamble was one of the most interesting articles which I have read for quite some time. It could easily serve as a general guide to basic amplifier design and pointing out some of the pitfalls of design and, of course, teaching one how to gain more useful information from the various technical data leaflets published by the manu-facturers of stereo equipment.

I particularly liked the point about distortion at low levels. I had, up till now, thought only that distortion at the high power levels to be of importance, and how it could be "swept-under-the-carpet" by plotting the graph with a linear scale instead

of a logarithmic one.

Of course, my impression of this preamble is purely subjective, not being connected in any way professionally with electronics, only in an amateur capacity, but still nevertheless, possessing a fairly good knowledge of basic valve and transistor theory through spare time reading of your own journal and various other publications. I only hope that next month's article can achieve the same level of interest; and I suspect that with the same author it will probably surpass it. As always, looking forward to next month's issue.

D. J. Whitaker, Bromley, Kent.

" GEMINI"

When the name P.E. Gemini was adopted for the Practical Electronics Stereo Amplifier, we overlooked the fact that Gemini was already associated with products of the Triplétone Manufacturing Co. This company has in fact used the name Gemini for a range of low cost amplifiers since 1959. The current version Tripletone Gemini Mk II is an all solid state design, previous models being valve designs. The output is 5W per channel and the retail price is £23 10s.

We hope there has been no confusion. In fact the P.E. Gemini and the Tripletone Gemini Mk II are quite dissimilar, both in specification and physical appearance.—Ed.

Beginning

Sir-I was interested to read Mr Bennett's letter (November 1970) and I appreciate the difficulty he and other beginners experience trying to find suitable books to guide them. Perhaps my own experience will be of interest to anyone finding it difficult to get started.

My interest in electronics began about 11 years ago at the age of At that time transistors were relatively expensive, but I found that old televisions and radios could be acquired for next to nothing, so naturally I began by building valve circuits. I visited my local public library, renounced the children's section forever, and looked instead at the mysterious books on electronics. Nearly all of these books explained how the various electronic components functioned but hardly any of them gave practical advice on making a circuit work.

It was around this time that I discovered that magazines were available describing practical circuits and they proved to be very useful. There was nearly always some simple project that I could build and the more complex designs could be left until later. I quickly found that the practical experience helped to make the theory more meaningful.

After a year or two I was building circuits to my own designs and some of the circuits behaved in the most extraordinary way. There were audio amplifiers which oscillated and audio oscillators which didn't. This was very frustrating at the time, but clearing the faults was a challenge and further knowledge was gained in Progressing from one this way. mistake to another may not sound exciting, but its amazing how much you learn. Touching the tags of a capacitor charged to 300 volts d.c. is a mistake that you learn not to repeat too often!

Transistors and small components bring electronics within the reach of anyone having a table and a few hours to spare. A 9 volt battery makes an excellent (and inexpensive) power supply and a surprising number of projects can be built around just two transistors.

My advice to anyone thinking of taking up electronics is quite simplehave a go! Read as many books and magazines as possible and spend as much time as you can trying out the circuits that are published.

A. M. Rudkin B.Sc.(Eng.), St. Albans, Herts.

Second string boost

Sir-Your reader Mr A. D. Jones (Readout, November 1970) refers to my article on guitar pick-up construction (September 1970) and points out that an important point was omitted. He is, of course, quite correct in his comment that the magnets should be mounted with like poles at the same end and I regret that this point was not included in the article.

The phenomenon known "second string boost" amongst guitar players, has for a long time been believed to be due to the particular character of the material, tension, etc. of the second string and it is supposed that this string vibrates with a slightly wider arc than the rest of the set, so producing a stronger

signal than its neighbours.

Mounting magnets with opposite poles adjacent will cause interaction between strings and certainly aggravate any tendency the second string may have to vibrate over a wider arc. Fortunately, modern technology has done much to eliminate this irritating feature and good makes of guitar strings do not usually produce this bugbear.

I have made quite a number of pick-ups in the manner described in the article, assuring that all magnets are the same way up, and all have produced an even response on all

strings.

L. F. Dickson, Salisbury, Wilts.

MEETINGS...

INSTITUTION OF **ELECTRICAL ENGINEERS**

LONDON

January 11, 5.30 p.m.

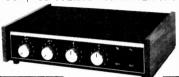
Amorphous Semiconductors, R. W. Brander, at Savoy Place. London, WC2R 0BL.

INSTITUTION OF ELECTRONIC AND RADIO ENGINEERS

LONDON

January 13, 6.0 p.m. Integrated Circuits for Colour Television, J. C. MacKellar, at Bedford Square, London, WCIB 3RG.

HARVERSONIC SUPER SOUND 10 - 10 STEREO AMPLIFIER KIT



NEW IMPROVED MODEL WITH HIGHER OUTPUT AND INCORPORATING HIGH QUALITY READY DRILLED PRINTED CIRCUIT BOARD FOR EASY CON-G HIGH PRINTED STRUCTION

A really first-class Hi-Fi Stereo Amplifier Kit. Uses 14 transistors including Silicon Transistors in the first five stages on each channel resulting in even lower noise level with improved sensitivity. Integrated pre-amp with Bass, Treble and two Volume Controls. Suitable for use with Ceramic or Crystal catridges. Output stage for any speakers from 5 to 15 ohms. Compact design, all parts supplied including drilled metal work, high quality ready drilled printed circuit board, attractive front panel, knobs, wire, solder, nuts, bolts—no extras to buy. Simple step by step instructions enable any constructor to build an amplifier to be proud of. Brief specification: Power output 14W r.u.s. per channel into 6 ohms. Frequency response + 34B 15-39,000Hz. Sensitivity better than 80mV into 1MQ. Full power bandwidth ± 34B 20-14,000Hz. Bass boost approx. to ± 12dB. Treble cut approx. to = 16dB. Negative feedback 18dB aver main amp. Power requirements 35V at 1-0 amp. Overall size—12° wide - 8° deep - 22° high. Pully detailed 7-page construction manual and parts list free with kit or send 3/h plus large 8.4.E.

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AMPLIFIER KIT. 210.10,0, P. & P. 3/-,
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220,100. Vost Free. Note: The above amplifier is suitable for feeding two mono sources into inputs (e.g. mike, radio, twin record decks, etc.) and will then provide mixing and fading facilities for medium powered Hi-Fi Discotheque use, etc.

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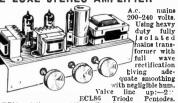
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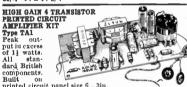
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8B5 The follow 8T1 8T2 8T3 8T4 8T5 8T6 8T7 8T7 8T7 8T7	300 lng are 50 40 40 30 30 40 40 40 30 30 30	8il. & Germ. Mixed Diodes all new and Pully Tested Transistor Packs. PNP Trans Mixed all coded Germ. NPN Germ. To-5 2G339 Eqyt. 2N1302 AC127. PNP Germ. AF 2G371B OC71-75 type PNP Germ. AF 2G371B OC71-75 type PNP Mein. VHF-2G407L= AF117= OC170 uncoded PNP Mein! Alloy VHF 800m/ca Trans. PNP/NPN Assorted Trans all coded. D1374 RF Trans Eqyt. ASV52 OC44/45 2N766 NPN Switching Sil. Trans. 2N760 NPN Switching Sil. Trans. 2N760 NPN Sil. Trans. BNY28 KPN Sil. Trans.	20 20 20 20 20 20 20 20 20 20 20 20 20 2
8B5 The follow 8T1 8T2 8T3 8T4 8T5 8T6 8T7 9T8 8T9 8T10 8T11	300 lng are 50 40 40 30 30 40 40 40 30 30 25	8il. & Germ. Mixed Diodes all new and Pully Tested Transistor Facks. PNF Trans Mixed all coded Germ. NFN Germ. To-5 2G339 Eqvt. 2N 1302 &C127 PNF Germ. AF 2G371B OC71-75 type. PNF Germ. AF 2G371B OC71-75 type. PNF Germ. VHF-2G407L= AF117=OC170 uncoded PNF Metal Alloy VHF 800m/og Trans. PNF/NFN Assorted Trans all coded. D1374 RF Trans Eqvt. ASV92 OC44/46 2N766 NFN Switching Sil. Trans. 2N768 NFN Sil. Trans. BSY28 NFN Sil. Trans. BSY28 NFN Sil. Trans.	20 20 20 20 20 20 20 20 20 20 20 20 20 2
8B5 The follow 8T1 8T2 8T3 8T4 8T5 8T6 8T7 8T7 8T8 8T7 8T8 8T7 8T10 8T11	300 ing are 50 40 40 30 40 40 40 30 25 25	8il. & Germ. Mixed Diodes all new and Fully Tested Transistor Packs. PNP Trans Mixed all coded Germ. NFN Germ. To-5 2G339 Eqvt. 2N1302 AC127. PNP Germ. AF 2G371B OC71-75 type. PNP Germ. VHF-2G407 L= AF117 = OC170 uncoded PNP Meial Alloy VHF 800m/sc Trans. PNP/NPN Assorted Trans all coded. D1374 RF Trans Eqvt. ASY32 OC44/46 2N766 NPN Switching Sil. Trans. 2N760 NPN Switching Sil. Trans. 2N3703 PNP Sil. Trans. 2N3703 PNP Sil. Trans. 2N3703 PNP Sil. Trans. 2N3703 PNP Sil. Trans.	20 20 20 20 20 20 20 20 20 20 20 20 20 2
8B5 The follow 8T1 8T2 8T3 8T4 8T5 8T6 8T7 9T8 8T9 8T10 8T11	300 lng are 50 40 40 30 30 40 40 40 30 30 25	8il. & Germ. Mixed Diodes all new and Pully Tested Transistor Facks. PNF Trans Mixed all coded Germ. NFN Germ. To-5 2G339 Eqvt. 2N 1302 &C127 PNF Germ. AF 2G371B OC71-75 type. PNF Germ. AF 2G371B OC71-75 type. PNF Germ. VHF-2G407L= AF117=OC170 uncoded PNF Metal Alloy VHF 800m/og Trans. PNF/NFN Assorted Trans all coded. D1374 RF Trans Eqvt. ASV92 OC44/46 2N766 NFN Switching Sil. Trans. 2N768 NFN Sil. Trans. BSY28 NFN Sil. Trans. BSY28 NFN Sil. Trans.	20 20 20 20 20 20 20 20 20 20 20 20 20 2

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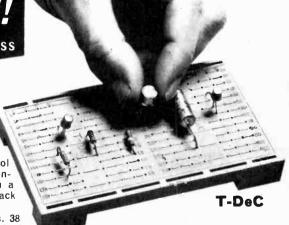
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TAG ENDS REF. No. G1/4 4/150V G1/5 8/275V G1/5A 32/350V G1/7 16/16/275V G1/12 40/450V	REF. No. 1/- G2/4	REF. No. 64/9 8/8/350V 1/- G4/10 350/25V 1/- G6/1A 3,000/15V 1/- G6/4 1,000/50V	2/- G6/7 100/275V 2/- G6/12 1,000/12V 3/- H1/9A 50/150V 6/- H3/5 250/150V	REF. No. 2/- H4/2 250/25V 2/- H4/7A 32/32/275V 1/- H4/12 500/50V 2/- H5/6 250/50V	1/6 2/6 2/- 2/-
PRINTED CIRCUIT, CA REF No. G1/6 16/32/450V G1/8A 16/350V G1/10 32/275V G1/14 16/275V G2/12 100/150V G2/12 100/150V G2/14 20/101/016/50V G3/16 100/250V G3/6 50/50/350V G3/6A 50/50/350V G3/16 100/250V G3/16 100/350V	NR TYPE 1/- G3/12 100/200/16/16/300 1/- G3/13 64/275V 1/- G4/3 250/150V 1/- G4/4 50/50/200V 1/- G4/8 40/40/450V 2/- G5/6 60/100/275V 2/- G5/6 100/100/255V 2/- G5/11 100/200/25V 2/- G5/11 100/200/25V 2/- G5/11 40/200/350V 2/- G5/11 40/200/350V 2/- G5/11 40/200/350V 2/- G5/11 40/40/350V	2/- G6/4A 1,000/25V 2/- G6/11 40/350V 2/- G6/12 250/25V 2/- H1/10A 10/10/350V 2/- H1/10A 10/10/350V 2/- H1/12 200/275V 2/- H1/13 250/150V 3/- H2/5 100/100/100/27 50V 2/6 H2/6 50/275V 2/6 H2/6 12/21 32/32/250V	REF. No. 1/6 H3/9 100/275V 2/6 H3/10 100/250V 1/- H3/13A 16/6/275V 2/- H4/3A 500/50V 2/- H4/4 40/6 (REV) 2/- H4/6 64/275V 2/- H4/7 32/3/50V 2/- H4/8 8/8/8/275V 2/- H4/8 8/8/8/275V 2/6 H4/10 10/10/425V 1/6 H4/13 400/250V 1/6 H5/1 150/200V	REF. No. 2/- H5/2 64/32/8/275V R5/2 14-5/3 500/25V (REV) 1/6 H5/8 32/32/275V R5/2 1/6 H5/8 32/32/275V R5/2 1/6 H5/19 100/100/150V R5/2 14-5/9 50/50/50/350V R5/2 14-6/4 150/150V R5/2 14-6/4 150/150V R5/2 14-6/4 11.000/20V R5/2 14-6/4 11.000/20V R5/2 14-6/14 11.000/20V R5/2 14-6/2 14	2/- 2/6 1/6 2/- 3/- 1/6 2/- 3/- 1/6 2/-
PRINTED CIRCUIT TY REF. No. G1/9 16/275V H2/7 10/25V H2/12A 350/12V H2/15 12.5/40V	PE, WIÑE LEADS REF. No. 1/6 H3/15 350/12V 1/- H5/1 150/30V 1/- H5/14A 1,100/10V 1/6 H6/5 150/25V	REF. No. 1/6 H6/10 300/9V 1/6 H6/11 300/15V 2/1 H6/14A 900/15V	REF. No. 1/- H7/12 65/15V 1/6 H7/15 100/16 2/- H8/1 1/350V	REF. No. 1/6 H8/4 5/15V 1/6 H8/8 20/9 1/- H8/9 40/15	1/-
SPECIAL 10/- PACKS. O	RDER IO PACKS AND WE	WILL INCLUDE SINCL	AIR AMPLIFIERS AND SPE	AKERS. Complete range in s	stock,
AN EXTRA ONE FREE RESISTORS, ½/½ watt assorted Wire-wound I to 3 watt 5 to 7 watt 10 watts Multi-tapped PAPER CONDENSERS	TRANSISTORS 100 10/- P.N.P. Untested b 20 10/- O.K.	All at VEROI ut mainly 50 10/- 3½in. × 3½in. × 50 10/- 5in. × 50 10/- Spot Fac	10 per cent discount on list.	n. × .15 5/6 3½in. × 3½in. × n. × .15 11/- 5in. × 2½in. × n. × .15 14/8 5in. × 3½in. × n. × .1 4/2 Tin. boards and Spot Face Cutter ,	.1 4/9 .1 4/7 .1 5/6

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MULLARD 'UNIFLEX' AMPLIFIERS
In our opinion these units are the best value for money ever offered. A complete stereo unit consisting of Control unit. Pre-amplifier, Two Main Amplifiers and Power Pack complete and ready for use—NO extra components to buy—yours for £15. (Normal retail price is £16.10s.)

GARRARD SP 25 UNITS also offered at a discount. Our price only £12.10s. Postage 9/-.

TRANSISTOR RADIOS

Once again we have a supply of these excellent radios which offer superb quality sound and excellent sensitivity. They are packed in a colourful presentation box complete with battery, earpiece and carrying case. Each one is guaranteed. You would expect to pay at least £5—but our price due to a bulk purchase is only 37/6. Many of our customers have made a considerable profit by buying these radios and re-selling them to friends!

RECORD PLAYER CARTRIDGES. Well below normal shop price! ACOS GP 67/2 15/-. (Mono). ACOS GP 91/3 20/-. (Compatible) ACOS GP 93/1 25/-. (Stereo) ACOS GP 94/- 30/-. (Stereo, Ceramic). ACOS GP 93/- DIAMOND 32/6. ACOS 94/1 DIAMOND 37/6.

TRANSISTORISED FLUORESCENT LIGHTS. 12 volt. All with Reverse Polarity protection. 8 watt 59/6. 15 watt 79/6. 13 watt 99/6. Postage 3/- per fitting. These can be sent on approval against full payment. Thousands already sold.

MULLARD POLYESTER CONDENSERS 1,000 pf., 1,200 pf., 1,800 pf., 2,200 pf., 400 volts, 3d, each. 25 per cent discount 100 any type. .15uf., .22uf., 27uf., 160 volts 6d. each. 25 per cent discount 100 any one type.

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1 watt to 50 watts. A large percentage of these are multi-tapped droppers for
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RECORDING TAPE BARGAIN! The very best British Made low-noise, high-quality Tape! Sin. Standard, 7/6. Long-play 9/-. Long-play 12/-. 7in. Standard 12/-. Long-play 16/3. We are getting a fantastic number of repeat orders for this tape. Might we suggest that you order now whilst we still have a good stock available?

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Thinking of learning another language? If so, apply to us for details of Linguaphone courses. We will GIVE you 65 worth of components of your selection when you purchase a language course"!!!!

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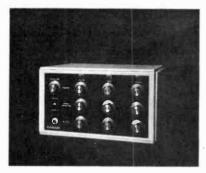
This fantastic brand new Astrad Auriga Russian 8 *COMPARE waveband portable radio, designed for world wide reception will probably make your present radio sense. If S PERFORMANCE with like a "crystal set"! It's far better than any other 392 6 18 RADIOS! battery and nosins use: We're almost giving the same and on the teem they have produced before away it only \$13.19.6 a mere fraction of even today's Russian miracle price; and nosins use: We're almost giving the one way and my state of the same and the same



Comparable with specifications costing around \$140, to be sold by others in 1. K. at \$279 as recently as Sept., '69. NOW—unbelievable but true—We are able to onler 500 only. BRAND SPAMKING NEW with writen 12 months guarantee, for only 49! gns., carriage, etc., 2 gns. You'll never appreciate the wonderful reproduction possible from your old stere or mone records until you listen to them on this superb Hi-Fi Audio Unit. It's a breath-taking experience! But wait. With the 6 Waveband Radio you can get thousands of transmissions from all never the world including ham radio, ships at sea, continental and international stationa. Amazing charity of tone and incredible station selection. Just read the magnificent specification of this compact fully transistorised, high quality sound reproduction equipment in matching unit form: Four-speed single Record Player deck, lightweight arm with "turnover" sapphire stylus and transparent deck cover, etc. (15/in 21m e9/in approx.). Two speaker Cabinets (sech 18/in-29/in-29/in approx.) cach containing 8in high efficiency unit and 3in cone tweeter. Fully transistorised Tuner/Amplifier (26 transistors; It diodes) (18/in-39/in-39/in approx.) incorporating VHF/FM, Long, Medium and 3 Short wavebands. All instantaneous Press Button top pacel controls. Push-pulloutput stage. Frontally controlled rotatable ferrite rod aerial, internal FM aerial, meter type tuning indicator, stereo balance control, separate variable bass and tratel control separate variable bass and tratel control separate variable bass and tratel of the control separate variable bass and tratel control separate variable bass and tratel of the strategy constructed—with beautiful high gloss wantup pulyester finish. Long negotiations and a final purchase-for cash made this incredible offer possible!

Please don't ask questions—just hurry! Only 500 will be luck! SPARES AND UK SERVICE CONSTANTLY AVAILABLE. Only 49; nn. plus carriage, crating, cost you clied to the supering decented by the present of the special control of the c Comparable with specifications costing around £140, to be sold by others in U.K. at £79 as recently as Sept., '69. NOW—unbelievable but true—We are able to

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The DABAR range of Kinetic Lighting Effects encompass a wide range of sophisticated colour pulsing and blending units, designed for discotheques, dance halls, restaurants, shop-window displays, etc. Basically 3 facilities are available: A sound to light Psychodelic lighting unit or colour organ. This unit connects directly into an amplifier unit or colour organ. Inis unit connects directly into an amplifier output or across loudspeakers (with negligible loading effect) and converts the audio frequency signals into a 3-colour light display; the colour depending on the frequency of the signal and the intensity of the light on the loudness of the audio source. Secondly a background of the light on the loudness of the audio source. Secondly a background control providing a variable minimum ambient lighting level if required and thirdly a colour blender providing a continuously and completely random slowly changing pattern of lights. This is fully automatic requiring no audio input and is controlled by 3 speed-control potentio-

requiring no audio input and it countries of their lkW of lighting per channel (3kW max per unit) or 3kW per channel (9kW max per unit). Latest full-wave Triac circuitry is used and radio-frequency interference filters are fitted in all units. Case size: 13in × 7in × 9in.

Audio Activated Colour organ incorporating background controls.

IkW per channel £50 3kW per channel £88

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IkW per channel £40 3kW per channel £78
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100 WATT AMPLIFIER

The latest addition to the DABAR range of equipment is a 4 channel integrated Preamplifier and Power amplifier, delivering a full 100W r.m.s. into a 4 ohm load. The amplifier has 4 inputs, each with their own gain control and a further master gain control, treble and bass control. A unique feature of this amplifier is in it's versatility of uses. It has been designed to accept plug-in pre-amp cards for each channel thus covering a wide range of applications. The basic unit is supplied with 4 input modules of your choice, but further modules may be purchased at moderate cost, thus enabling with the minimum of time purchased at moderate cost, thus enabling, with the minimum of time, effort and outlay the conversion of any one or all channels to accept an

enort and outlay the conversion of any one or all channels to accept an entirely different input. Modules available include: P/U cartridges—crystal, ceramic, and magnetic (equalised R.I.A.A.); Microphones— $30-60\Omega$, $300-600\Omega$ and $50k\Omega$ types. Guitar and two modules for high impedance outputs from tape, tuner, etc., 100mV f.e.t. and 1V input. Case size: $13in \times 7in \times 9in$.

Price complete with 4 modules £109.0.0 Extra modules £5.0.0 each.

Please send J/- in stamps for full descriptive literature.

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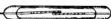
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