

# VIDEO SWEEP GENERATOR

WA-21A

RADIO CORPORATION OF AMERICA

Camden, New Jersey, U. S. A.



#### ADDENDA TO IB-4021-2



**TO INCLUDE** 

## **TYPE WA-21B VIDEO SWEEP GENERATOR**

#### MI-30021-B

Information pertaining to the Type WA-21A Video Sweep Generator included in instruction book IB-4021-2 also applies to the Type WA-21B Equipment, with the following changes:

1. Under Tube Complement, opposite Video Amplifier (page 4) change "1 RCA-6C4" to "1 RCA-6J6." Also, opposite Video Amplifier change "2 RCA-6AG5" to "2 RCA-6AH6."

2. Under Circuit Description, Mixer (page 9) substitute "RCA-6J6" for "RCA-6C4."

3. In Figure 13 (page 12) change the symbol

number of the Adjustment marked "L21," to "L28," and delete adjustment "C88."

4. Delete the entire section headed Adjustment of the Modulation Capacitor, including figures 15, 16 and 17 (pages 13, 14 and 15).

5. Under Aligning the Video Amplifier, on page 15, change "L21" to "L28."

6. Use the following *Replacement Parts List* and *Schematic Diagram* instead of those in the instruction book.

7. Refer to Connection Diagram in this addenda for wiring and parts location.

Symbol No.	Description	Stock No.	Symbol No.	Description	Stock No.
C1	Capacitor, fixed, mica, 2700 mmf		C30	Same as C4	
C2 C2	$\pm 10\%$ , 500 v	39662	C31	Same as C8	
C2, C3	Capacitor, fixed, mica, 56 mmf	20(22	C32	Same as C4	20/0/
C4	$\pm 10\%$ , 500 v Capacitor, ceramic, high "K"	39622	C33 C34, C35	Capacitor, fixed, mica, 12 mmf	39606
Ci	type, 1000 mmf $\pm 20\%$ , 300 v	71929	C36	Same as C28 Same as C2	
C5	Same as C1	/1929	C37, C38	Same as C4	
Čć	Capacitor, fixed, mica, 3300 mmf		C39	Capacitor, paper oil filled, 2	
	$\pm 20\%$ , 500 v	39664	0.57	sections, .5 mf each $\pm 15\%$ ,	
C7	Same as C4	57001		400 v	71673
Č8	Capacitor, fixed, mica, 5600 mmf	39670	C40	Same as C4	
C9	Capacitor, fixed, mica, 82 mmf		C41	Capacitor, ceramic, non-insulated,	
	$\pm 10\%$ , 300 v	55328	_	$9 \text{ mmf} \pm 10\%, 500 \text{ v}$	55571
C10	Capacitor, fixed, mica, 10 mmf		C42	Capacitor, fixed, mica, 200 mmf	
	$\pm 10\%$ , 500 v	75245		±5%, 500 v	39635
C11, C12	Same as C4		C43	Same as C4	
C13	Capacitor, fixed, mica, 22 mmf		C44, C51	Same as C39; C44 section 1, C51	
	$\pm 10\%$ , 500 v	39612		section 2	
C15	Capacitor, fixed, mica, 1000 mmf		C45	Same as C39	
	$\pm 10\%$ , 500 v.	39652	C46	Same as C39	
C16	Same as C1		C47, C48	Same as C4	
C17, C18	Same as C4		C49	Same as C42	
C19	Capacitor, molded head lead type, 4.7 mmf, 500 v	54402	C50 C51	Same as C41 See C44	
C20	Same as C4	54402	C52, C61	Same as C39; C52 section 2, C61	
C20 C21	Capacitor, ceramic, high "K"		0,2,001	section 1	
C21	type, 1500 mmf $\pm 20\%$ , 300 v	58244	C53, C54,	Section	
C22	Same as C13	<i>JOL 11</i>	C55	Same as C4	
Č23, C24	Same as C4		C56	Same as C39	
C25	Same as C10		C57	Same as C4	
Č26	Same as C4		C58	Same as C41	
C27	Capacitor, variable, ceramic, 5-20		C59	Same as C42	
	mmf	55301	C60	Same as C4	
C28	Capacitor, fixed, mica, 100 mmf		C61	See C52	
	$\pm 10\%$ , 300 v	39628	C62	Same as C1	
C29	Capacitor, variable, 7-100 mmf	18384	C63	Same as C4	

#### **PARTS LIST**

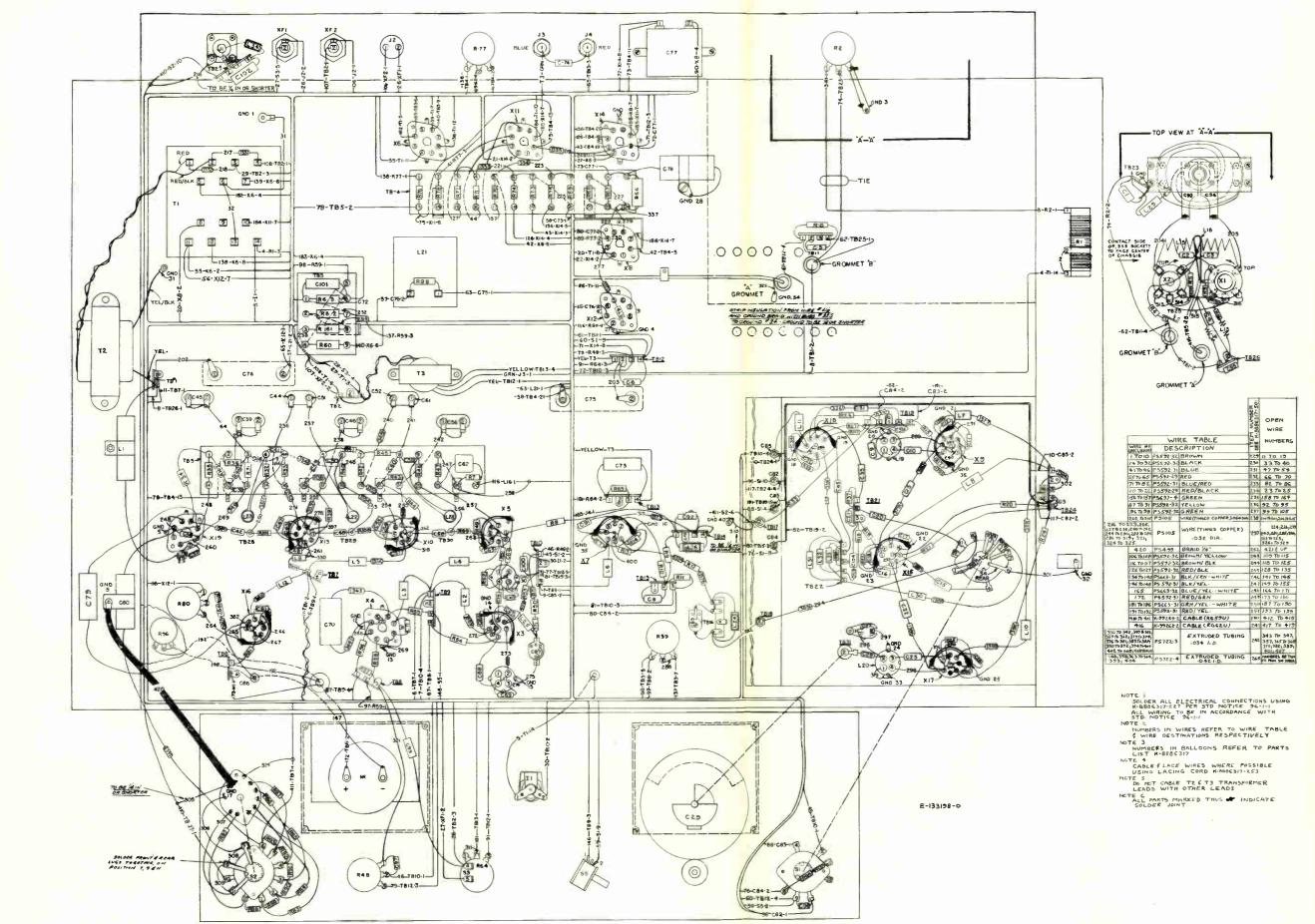
IB-4021-2A

#### PARTS LIST (Cont.)

Symbol No.	Description	Stock No.
C64	Capacitor, fixed, mica, 33 mmf $\pm 10\%$ , 300 v	39616
C66, C67 C68	Same as C4 Capacitor, fixed, mica, 39 mmf $\pm 10\%$ , 300 v	39618
C69 C70 C71	Same as C13 Capacitor, paper tubular, .25 mf $\pm 20\%$ , 400 v	54145
C72	Same as C4 Capacitor, paper tubular, .025 mf $\pm 10\%$ , 400 v	70612
C73 C74	Same as C70 Capacitor, paper tubular, .1 mf $\pm$ 20%, 400 v	73551
C75, C76 C77	Capacitor, paper, 10 mf, 600 v Capacitor, mineral oil treated, 1 mf $\pm 10\%$ , 400 v	52734 45807
C78	Capacitor, dry electrolytic, 20 mf $\pm 10\%$ , 450 v	56243
C79, C80 C81 C82 to C86	Same as C70 Same as C4 Capacitor, feed-thru, 500 mmf	
inc. C87	±20%, 350 v Capacitor, variable trimmer, 2-6 mmf	55300 54678
C89 C91	Capacitor, variable trimmer, 5-20 mmf Capacitor, fixed, mica, 4700 mmf	64689
C92	$\pm$ 5%, 500 v Capacitor, variable trimmer, 5-50	53112
C95, C96, C97	mmf Same as C4	54681
C99 C100 C101	Same as C4 Same as C91 Capacitor, paper tubular, .005 mf	
C102	$\pm 10\%$ , 1000 v Capacitor, molded, paper, .47 mf $\pm 20\%$ , 200 v	70648 73787
C103 C104 C105, C106	$\pm$ 20%, 200 v Same as C41 Same as C33 Same as C13	
F1, F2 J1	Fuse, 3 amp, cartridge type, glass body Plug, output, amphenol, single	10907
J2 J3 J4	contact, chassis mtg. Connector, motor connector base- Jack, pin jack (Blue) Jack, pin jack (Red)	51800 47594 55239 55238
L1 L2, L3 L4, L5 L6, L7, L8 L9	Not Stocked Coil, filament choke Coil, filament choke Same as L2 Same as L4	95850 95851
L10, L11, L12	Same as L2 Not Stocked	
L13, L14 L15, L16 L17	Not Stocked Transformer, sweep heterodyne	
L18 L19	freq. control adjusting Transformer, doubler Transformer, fixed marker oscil-	55563 55564
L20	lator freq. control inductor Transformer, variable marker osc., freq. control inductor	55565 55566
L21 L22, L23 L24	Coil, choke Coil, peaking Coil, R.F. choke	48554 94832 54683
L25 L26 L27	Coil, R.F. choke Same as L24 Capacitor, vibrator capacitor as-	55570
L28	sem. (Includes C93, C94) Same as L22	59756

SumbalNa		Stock
Symbol No.	Description	No.
L29	Coil, R.F. choke, 4.7 micro. hen.	76510
L30, L31,	Coil, R.F. choke, 10 micro. hen.	95849
L32		
L33	Same as L29	0.00 (0
L34 M1	Coil, R.F. choke, .47 micro. hen. Meter, D.C., 0 to 200 micro.	95848
MII	amps. D.C. calibrated 0 to .6	
-	volts	55557
R1	Resistor, adjustable, wire wound, 15 ohms	48547
R2	Resistor, variable, 6 ohms, 25 w	55567
R3	Resistor, fixed, composition,	
R4	12,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	30436
<b>N</b> 4	Resistor, fixed, composition, 680 ohms, ±10%, 1 w	19233
R5	Same as R3	
R6	Resistor, fixed, composition,	60000 <b>-</b>
R7	2700 ohms, 2 w Resistor, fixed, composition,	523227
	100 ohms, $\pm 10\% \frac{1}{2}$ w	502110
R8	Resistor, fixed, composition, 4700 ohms, ±10%, ½ w	502247
R9	Resistor, fixed, composition,	502247
<b>D</b> 10	820 ohms, ±10% 1 w	68025
R10 R11	Same as R3 Resistor, fixed, composition,	
	10,000 ohms, ±10%, 1 w	71914
R12	Resistor, fixed, composition,	100/0
R13	680 ohms, $\pm 10\%$ , $\frac{1}{2}$ w Resistor, fixed, composition,	12262
-	220 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	5201
R14	Resistor, fixed, composition,	502/10
R15	100,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w Resistor, fixed, composition,	502410
	Resistor, fixed, composition, 82,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	8064
R17	Resistor, fixed, composition, 330,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	14983
R18, R19	Resistor, fixed, composition,	11/05
R20	47,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w Resistor, fixed, composition,	30787
R20	1000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	34766
R21	Resistor, fixed, composition,	602222
R22	22,000 ohms Resistor, fixed, composition,	503322
	270 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	30929
R24	Resistor, fixed, composition, 120 ohms, ±10%, ½ w	502112
R25	Same as $R17$	J02112
R26	Same as R18	
R27	Resistor, fixed, composition,	20007
R28	68,000 ohms, $\pm 10\%$ , 1 w Resistor, fixed, composition,	38897
-	330 ohms, ±10%, ½ w	8063
R29	Resistor, fixed, composition, 680 ohms, ±10%, ½ w	12262
R30	Resistor, fixed, composition,	
•	3900 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	30694
R31	Resistor, fixed, composition, 100 ohms, ±10%, 1 w	31215
R32	Resistor, fixed, composition,	
D 2 2	680 ohms, ±5%, 1 w Resistor, fixed, composition,	512168
R33	$39,000 \text{ ohms}, \pm 10\%, \frac{1}{2} \text{ w}$	30147
R34	Same as R14	
R35	Resistor, fixed, composition, 180 ohms, ±10%, ½ w	502118
R36	Same as $R_{31}$	502110
R37	Same as R15	
R38	Resistor, fixed, composition,	26762
R39	68 ohms, ±10%, ½ w Resistor, fixed, composition,	34763
NJ7	180,000 ohms, $\pm 10\%$ , $\frac{1}{2}$ w	11959
	100,000 0000, 210/0, 72 **	





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## VIDEO SWEEP GENERATOR

WA-21A

## INSTRUCTIONS

## RADIO CORPORATION OF AMERICA Camden, New Jersey, U. S. A.

IB-4021-2

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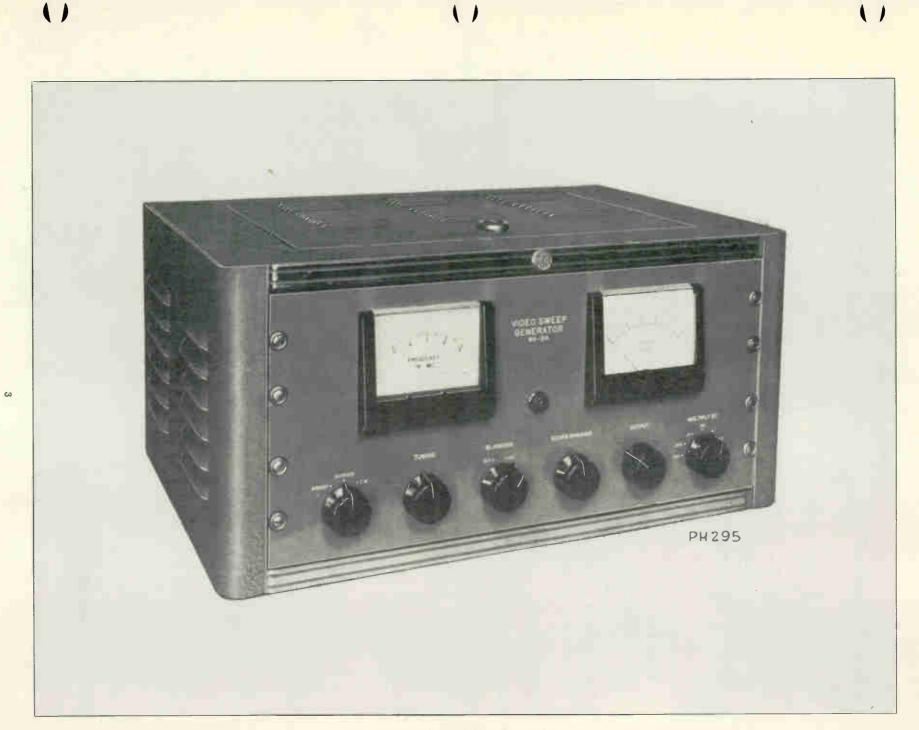


Figure 1-WA-21A Video Sweep Generator

## TECHNICAL SUMMARY

## Electrical Characteristics

#### Sweep Output

Swept	Frequency	Range	. 100 kc to 10 mc
Outpu	t Voltage .		one max. volts rms
Outpu	t Impedanc		ohm terminated line
Outpu	t Amplitud	e Characteristics flat within one db of mid-frequency output fr	om 100 kc to 10 mc
Distor	tion and Sp	ùrious Output	total less than 5%

#### C-W Output

Frequency Range	continuously variable from 100 kc to 10 mc
Dial Scale Error	. less than 2% or 50 kc (whichever is greater) at any frequency setting
Output Voltage	
Output Impedance	
Output Amplitude Characteristics	essentially flat from 100 kc to 10 mc
Distortion and Spurious Output	
Power Requirements	105-125 volts, 50-60 cycles, 180 watts
Fuse Protection	

#### Tube Complement

2 R	CA-955	sweep oscillator
1 R	СА-6ВА6	oscillator-doubler
1 R	CA-6SQ7	blanking tube
1 R	CA-6C4	sweep mixer
2 R	CA-6AG5	video amplifier
1 R	CA-6AG7	video output
1 R	CA-6BE6	marker mixer
1 R	CA-6AG5	fixed marker oscillator
1 R	CA-6AG5	variable marker oscillator
	CA-6AG5	
1 6.		c-w mixer
16. 1R	AS6	c-w mixer buffer amplifier
16. 1R 1R	AS6	c-w mixer buffer amplifier rectifier
1 6. 1 R 1 R 1 R 1 R	AS6 CA-6AG5 .CA-5U4G .CA-OD-3 CA-6SJ7	c-w mixer buffer amplifier rectifier voltage regulator d-c amplifier
1 6. 1 R 1 R 1 R 1 R	AS6 CA-6AG5 .CA-5U4G .CA-OD-3	c-w mixer buffer amplifier rectifier voltage regulator d-c amplifier

## Mechanical Characteristics

Dimensions	10 <mark>1/2</mark> in.	high,	22 in.	wide, 15 in. deep
Weight	• • • • • • •	• • • • • •	80	pounds (approx.)

#### GENERAL DESCRIPTION

Designed to facilitate rapid testing of video-frequency networks and routine maintenance checking of television transmitter and terminal equipment, the Video Sweep Generator, Type WA-21A, in connection with a suitable detector and a general-purpose oscilloscope, permits continuous observation of the gain-versusfrequency characteristic of the circuit under test. It also provides sine-wave signals tuneable from 100 kc to 10 mc for point-by-point or steady-state tests.

The section of the circuit which produces the continuous-wave signal is separate from that which generates the swept signal, thus allowing the c-w signal to serve as a calibrating marker signal on the oscilloscope trace.

The swept signal covers the range from 100 kc to 10 mc at the rate of 60 complete excursions per second. Over this range, the output is uniform within one decibel of the mid-range output. Stability of the sweep signal is insured by the use of a mechanically-operated modulating capacitor. Distortion and spurious signal outputs are less than five percent of the fundamental output voltage.

A blanking circuit is provided to cut off the signal during the return portion of the oscilloscope trace (from 10 mc to 100 kc) if so desired, thus furnishing a zero-output base line on the trace. If the marker is utilized, a clearly visible "pip" appears on the trace, the frequency at which this mark occurs being read directly from the marker tuning dial. The full marker frequency range from 100 kc to 10 mc is covered by a single dial scale having an error of less than 2% or 50 kc (whichever is greater).

A method of developing the marker pip is employed which minimizes spurious responses and makes the pip uniform in size over the entire tuning range.

A panel switch permits selection of sweep alone, sweep plus the marker signal, or the c-w signal alone. The c-w output is supplied through the same coaxial line as the sweep signal.

A source of 60-cycle voltage, adjustable in phase, is provided for connection to the oscilloscope horizontal deflection terminals for convenience in case this feature is not already included in the oscilloscope.

A metering circuit is provided to indicate the rms value of the output signal voltage. This circuit consists of a diode-type voltmeter used in connection with a resistive step attenuator. Continuous coverage of the output voltage range from one volt to 0.5 millivolt is achieved through the use of a fine output control and the step attenuator.

The instrument is designed for table-top mounting and is normally furnished in a table-model cabinet; however, the unit is suitable for rack mounting if desired. A dust-cover is provided which serves to protect the top of the chassis when relay-rack mounting is used. This cover is superfluous for cabinet mounting, and may be discarded to afford added ventilation. Standard relay-rack mounting slots are furnished on the front panel.

#### INSTALLATION

After carefully unpacking the instrument, prepare it for operation as follows:

1. Remove the eight screws from the front panel and withdraw the panel-and-chassis assembly from the case.

2. Remove the screws on the rear of the instrument which hold the dust-cover in place, then lift off the dust-cover.

3. Inspect the tubes, and see that they are all firmly seated in their sockets.

4. Turn the SCOPE PHASING control fully counterclockwise. See that the two fuses on the rear apron are marked three amperes and are in good condition. Insert the power cord supplied with the instrument into the receptacle on the rear apron, and plug the other end of the cord into an a-c outlet supplying 105-125 volts, 50-60 cycles.

5. Adjust the VOLTAGE REGULATOR control as described in the Maintenance section of this book.

6. If cabinet mounting is preferred, store the dustcover for possible future use, and replace the paneland-chassis assembly within the cabinet.

7. If relay-rack mounting is preferred, replace the dust-cover, and store the cabinet for future use. Bolt the panel-and-chassis assembly to the relay rack with 10/32 nuts and bolts.

8. Insert the output cable into its receptacle on the rear apron. The instrument is now ready for use.

#### FUNCTION OF OPERATING CONTROLS

SWEEP-MARKER-CW Switch—This switch selects the mode of operation of the instrument. In the "SWEEP" position, the switch makes a sweep signal available at the end of the output cable. When the switch is on "MARKER," a sweep signal with a superimposed marker "pip" is present at the end of the cable. A c-w signal is made available when the switch is on "CW."

TUNING Control — This control adjusts the frequency of the marker signal or of the c-w signal to the value shown on the dial scale.

BLANKING Switch — When this switch is in the "OFF" position, a conventional double trace is made to appear on the oscilloscope screen. A reference baseline representing zero level is caused to appear on the screen when the switch is in the "ON" position. The blanking is timed to occur during the return sweep interval (from 10 mc to 100 kc).

SCOPE PHASING Control—Power is applied to the instrument when this control is rotated clockwise. Turning the control to its extreme counterclockwise position turns the instrument off. As the control is rotated, the phase of the horizontal deflection voltage available at the SCOPE PHASING terminals on the back apron is varied.

OUTPUT Control—This control is used as a vernier output level adjustment.

MULTIPLY BY Switch—Rough adjustments in output level are made with this switch. When the BLANKING switch is set to "OFF," the setting of the MULTIPLY BY switch times the panel-meter reading equals the rms value of the output voltage.

#### OPERATION

Initial Adjustments — The Video Sweep Generator, a suitable oscilloscope, and a detector should be set up as shown in Figure 2.

The HORIZONTAL DEFLECTION terminals on the WA-21A should be connected to the horizontal deflection terminals of the oscilloscope. The detector input should be connected to the WA-21A output, and the output of the detector should be connected to the vertical terminals of the oscilloscope. Both the detector output circuit and the oscilloscope should have good 60-cycle square-wave response. A suitable detector, used in forming the curves shown in this book, is shown in Figure 2.

Turn the instrument on by rotating the SCOPE PHASE control clockwise. Place the BLANKING switch in the OFF position, and the SWEEP-MARKER-CW switch on SWEEP. A pattern similar to that shown in Figure 3A should appear on the screen. The two notches represent zero frequency, and if they are displaced, the SCOPE PHASE control should be rotated until the notches are superimposed, as shown in Figure 3B. The reduced output at fre-

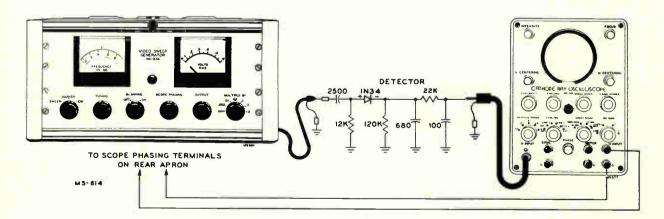


Figure 2-Setup for Initial Adjustment

quencies approaching zero is characteristic of the peak rectifying detector used in making the curves and is partially due to the limited low-frequency response of the instrument.

А 7 СV-59А В

Figure 3-Response Curves with Blanking Off

Throw the BLANKING switch to the "ON" position. A pattern similar to that shown in Figure 4 should be visible on the oscilloscope screen. If this pattern is obtained, the Video Sweep Generator is ready for testing a wide-band video amplifier or other device as covered in the section on Applications.

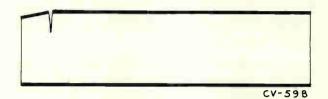


Figure 4-Response Curve with Blanking On

#### APPLICATIONS

Testing and Adjusting Video Amplifiers — The test setup is shown in Figure 5 and the general procedure follows:

1. Make the preliminary adjustments on the Video Sweep Generator as described under Operation. Adjust the controls on the oscilloscope for a trace of the desired height, width, and brilliance.

2. When the video amplifier is connected between the WA-21A and the detector, a trace representing the frequency-vs-amplitude characteristic of the amplifier will be seen on the oscilloscope. A baseline, representing zero response, may be obtained by rotating the BLANKING switch to "ON." The frequency of any point on the response may be determined by setting the SWEEP-MARKER-CW switch on "MARKER," and adjusting the TUNING control until the marker pip rests on the point under consideration. The frequency is then read from the tuning dial scale. 3. Adjust the peaking coils in the video amplifier for a response of the desired shape.

When testing video amplifiers, certain precautions must be observed to avoid erroneous results. One improper indication of the response of a video amplifier may be obtained when the output of the WA-21A is too great, causing limiting in the video amplifier. Figure 6 shows the effect obtained with this condition. The actual response is shown at "A"; but when the output of the WA-21A is of such magnitude as to cause limiting in the video amplifier, the curve of "B" is obtained. To avoid any confusion, the OUTPUT and MULTIPLY-BY controls on the Video Sweep Generator should be initially set for low output, then gradually advanced. If the shape of the response curve changes as the WA-21A output is increased, then limiting is indicated, and the OUTPUT control should be backed off.

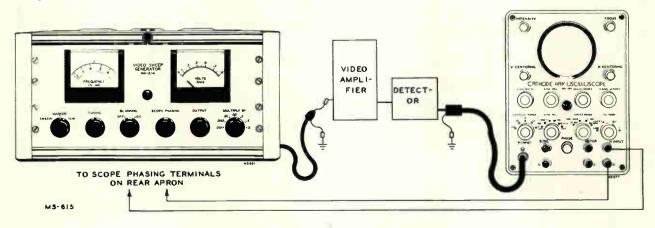


Figure 5---Video Amplifier Test Setup

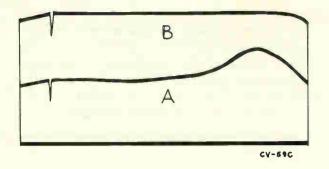
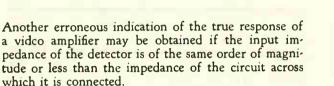


Figure 6-Effect of Limiting on Response



If the particular section of the amplifier where the detector is connected is of relatively low impedance, such as the output of a cathode follower, then no special precautions are necessary; however, if the detector is connected across a high-impedance interstage coupling network, the additional capacity loading will cause the observed response to be inaccurate. This effect must be taken into consideration when interpreting response curves.

One way to check the response of a high-impedance circuit feeding the grid of a following stage is shown in Figure 7. The 200-ohm resistor across the plate load of V2 minimizes the capacity loading effect of the detector on the preceding stage (V1). In this method V2 acts essentially as a buffer, and the response observed is the response of the circuit only up to the grid of V2. This method does not show the response of the stage to which the detector and the loading resistor are connected; i. e., in the circuit of Figure 7, the effect of the peaking coils in the plate circuit of V2 is not shown.

If the oscilloscope used to observe the response of the circuit under test utilizes a wide-band video amplifier, then it is not necessary to employ a detector in the test setup. In this case, the output of the video amplifier is fed directly to the vertical amplifier of the oscilloscope. A trace obtained using this method is shown in Figure 8. The successful employment of this method requires that the frequency response of the

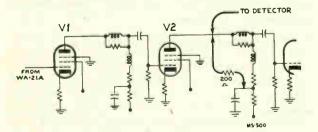


Figure 7—Temporary Circuit Modification of Video Amplifier

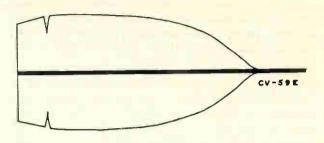


Figure 8-Response Without Detector

oscilloscope vertical amplifier be flat over a range at least as great as the entire band passed by the video amplifier under test.

When aligning high-gain video amplifiers consisting of a number of stages in cascade, with the Video Sweep Generator connected to the input, and the oscilloscope connected to the output, undesirable coupling between the input and the output through the power-supply lines may be experienced. This coupling usually evidences itself as a change in the shape of the response curve when the cases or the cables of the equipment used in the test setup are touched. To minimize this effect, place a large sheet of metal under all of the equipment, and connect or bypass all units to this common metal plate.

Testing and Adjusting Filters and Frequency-Selective Networks — The test setup is the same as that shown for adjusting video amplifiers, except that the filter under investigation replaces the video amplifier of Figure 5. Different frequencies on the response trace can be identified by employing the marker. The filter variables are adjusted to give the required response as observed on the oscilloscope.

If the response of a relatively narrow-band filter is to be examined in detail, then the swept bandwidth of the WA-21A can be reduced. Figure 9 shows the effect of observing the response of a narrow-band circuit on the 0-10 mc sweep and on a sweep of reduced bandwidth. The Video Sweep Generator can be adjusted to sweep the required band of frequencies

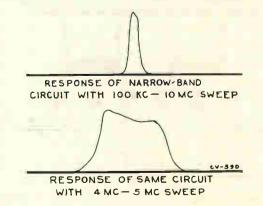


Figure 9—Narrow-Band Response Curves

according to the instructions given in the Maintenance section of this book. For examining the characteristics of very narrow band circuits, such as broadcastreceiver i f transformers, this method is not recommended.

When observing the frequency characteristic of a network having a sharp cutoff or steep sides in its response, a detector with good transient response must be used. Filters of different characteristics will require detectors of different design; i. e., a circuit having a high output impedance will require a detector that may be unsuitable for use with a circuit having sharp-cutoff characteristics.

In certain cases where a sharp-cutoff circuit is designed to operate into a relatively high impedance, the detector may have to be isolated from the circuit under investigation by means of a buffer stage, such as a cathode follower. In any case, a circuit consisting of a cathode follower feeding a detector having good transient response represents probably the most uni-

The WA-21A Video Sweep Generator circuit is shown schematically in Figure 18 and in functional block form in Figure 11.

Sweep Oscillator — Two RCA type 955 triodes (V1, V2) are employed in a push-pull oscillator circuit which is frequency-modulated over an approximate range of 70.80 mc by a vibrating capacitor (C93, C94) of special design. One plate of this capacitor is mechanically coupled to an actuating coil (L27) which is excited from the 60-cycle power source. The frequency deviation produced by the modulation capacitor is dependent upon the amplitude of its vibration, which is a function of the amount of excitation applied to the actuating coil. The excitation, and hence the frequency deviation, is adjusted by means of the SWEEP WIDTH CONTROL (R2). For CW operation of the instrument, the oscillator plate voltage is removed.

Fixed Oscillator-Doubler — This circuit generates a fixed r-f signal of 70 mc which is heterodyned with the frequency-modulated signal provided by the sweep oscillator to form the final 0-10 mc video-frequency sweep. An RCA-6BA6 (V3) oscillates at 35 mc in an electron-coupled Hartley circuit. Frequency doubling is accomplished in the plate circuit of the tube since L18, the plate load, is tuned to 70 mc. A variable resistor (R48, OUTPUT control) adjusts the output of the oscillator-doubler by varying its plate voltage. Since varying the amplitude of the 70-mc signal varies the efficiency with which this signal is mixed with the sweep-oscillator signal, the OUTPUT control governs the amplitude of the 0-10 mc sweep signal fed to the video amplifier.

Mixer — An RCA-6C4 triode is utilized as a mixer. The 70-mc signal is injected into the grid of the

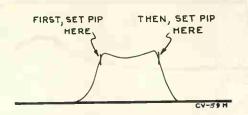


Figure 10-Determining Bandwidth

versally adaptable device for use with the Video Sweep Generator.

Determining Bandwidth of Frequency-Selective Circuits — The circuit under test is first set up in the usual manner so that its frequency response is displayed on the oscilloscope. The marker pip is then set first on one side of the response, and then on the other, as shown in Figure 10. The difference between the two frequencies, as read from the tuning-dial scale, is the bandwidth of the circuit under test.

#### CIRCUIT DESCRIPTION

mixer; the 70-80 mc signal generated by the sweep oscillator is fed in at the cathode. As a result of the heterodyne action occurring in the mixer tube, a video-frequency signal which is constantly swept 60 times a second over a range from 0 to 10 mc is developed in the mixer plate circuit. This sweepfrequency signal is coupled to the grid of the first video amplifier (V10).

Video Amplifier — A conventional two-stage, wideband video amplifier is employed in the instrument. The response characteristics of the amplifier are such that the output is flat within one db from 100 kc to 10 mc.

Video Output Stage — The output of the video amplifier is fed to the grid of an RCA-6AG7 (V19) connected as a cathode follower. This circuit, by virtue of its low internal impedance, permits the use of a low-impedance output and its attendant relative freedom from capacity-loading effects. Sweep voltage from the cathode of the video-output tube is fed to the output jack (J1) through a ladder-type attenuator (R82 to R93). This attenuator (MULTIPLY BY switch) is used to vary the output of the instrument in fixed steps. The coaxial cable which plugs into the output jack is fitted with an internal 75-ohm resistive termination.

Voltmeter and Voltmeter Diode — A sample of the video output is fed to the voltmeter diode just ahead of the attenuator circuit. The a-c axis of the output voltage is shifted in a negative direction by the diode (V16), which functions as a d-c restorer. The meter reading is thus made proportional to the peak value of the voltage fed to the attenuator. A variable resistor (R96) is used to calibrate the meter in rms values. The zero-signal meter current due to diode contact

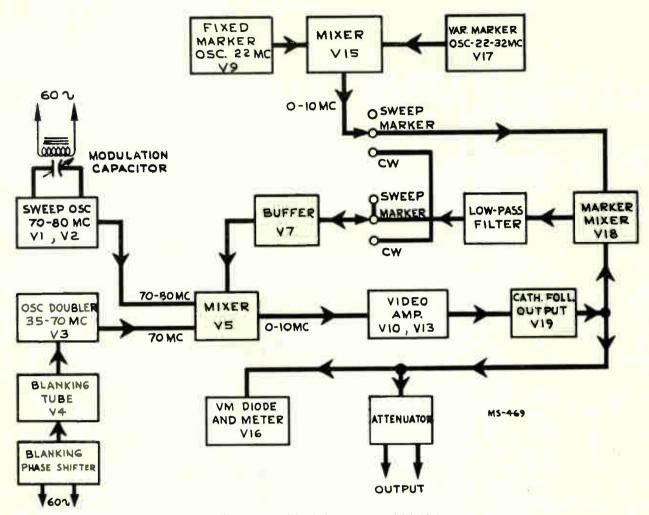


Figure 11-Block Diagram of WA-21A

potential is balanced out with the METER ZERO control (R80). The time-constant of the diode circuit is not sufficiently large to provide accurate readings when the blanking is on.

Marker and C-W Signal Generator — The marker generator comprises a 22-mc oscillator, a variablefrequency oscillator which can be tuned to any frequency between 22 and 32 mc, and a mixer which combines the two oscillator signals to produce the marker signal.

The 22-mc oscillator is an RCA-6AG5 (V9) in a conventional Hartley circuit. The variable-frequency oscillator (V17) is similar to the 22-mc oscillator except that its frequency can be tuned over the 22-32 mc range by means of the front-panel TUNING control (C29). Signals from the two oscillators are combined in the mixer (V15) to form, at the mixer plate, a marker signal which can be set anywhere between zero beat and 10 mc with the TUNING control.

When a marker signal is not desired, plate voltage for all tubes in the marker generator (V9, V15, V17) is removed by setting the front-panel selector switch (S1, S4) to "SWEEP."

Marker Mixer — This tube (V18) combines the marker signal with a sample of the video sweep output obtained from the video output tube (V19). The output of the marker mixer contains a beat between the marker frequency and the swept video frequency. This beat is fed to a low-pass filter, which passes only the low-frequency components of the beat. Since the low-frequency components are present only for an instant while the swept video frequency is sweeping through the marker frequency, the output of the lowpass filter contains a signal only during this instant. This signal is fed to the buffer amplifier through the selector switch (S4, front).

Buffer Amplifier — The output of the low-pass filter is received and amplified by an RCA-6AG5 (V7). Since the plate of this tube is conductively coupled to the plate of the mixer (V5), the 100 kc-10 mc sweep output of the mixer is amplitude modulated by the signal fed to the buffer grid. This amplitude modulation of the swept video voltage as it sweeps through the marker frequency is observed on the cathode-ray tube trace as a marker "pip."

C-W Output — When the selector switch (S1, S4)is on "CW," the sweep oscillator (V1, V2) plate voltage is removed and the output of the marker mixer (V18) is disconnected; therefore, no video sweep voltage is available from the instrument. The marker generator (V9, V15, V17), however, still functions as a variable-frequency source. Its output is switched to the buffer amplifier, which feeds the c-w signal to the video amplifier. The output voltage is controlled by the MULTIPLY BY switch and the OUTPUT control (R48), which varies the plate voltage of the 22-mc oscillator (V9).

Blanking Generator — When the BLANKING switch is turned to "ON," plate voltage is applied to the blanking tube (V4). The tube amplifies and clips the 60-cycle voltage which is applied to its grid through

Tube Replacement — Most of the troubles which develop in the instrument can be attributed to vacuum tubes which have reached the end of their useful lives, or which have developed defects. All tubes should be tested periodically. Since all tube sockets are marked with the tube type numbers, removal and replacement of tubes is a relatively simple matter. To obtain access to the tubes, lift the lid of the case. The two RCA 955 acorn tubes are located in a shielded compartment at the rear left-hand side of the chassis. Access to these tubes may be obtained by removing the compartment cover, which is held in place with six screws.

Sweep Output Bandwidth Adjustment — Normally, the instrument is preset at the factory for a swept bandwidth of 10 megacycles. To obtain this bandwidth, the sweep oscillator is adjusted to sweep a band of frequencies from 70 to 80 mc (approx.) and the oscillator doubler is adjusted to produce a c-w output of approximately 70 mc. Figure 12 illustrates the pattern of the output signal as seen on an oscilloscope when the instrument is properly adjusted for a 10-mc bandwidth. If the 10-mc marker does not fall near the right-hand side of the trace, or if the zero-frequency beat does not fall near the left-hand side of the trace, the following adjustments should be made:

1. Connect the SCOPE PHASING terminals on the rear apron of the WA-21A to the horizontal terminals of an oscilloscope.

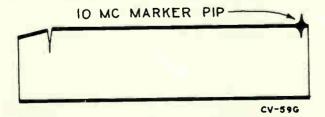


Figure 12-Response Curve for 10-Mc Bandwidth

the phase-shifting network R59 and C72. The square wave on the plate of the tube is fed to a d-c restorer (diodes of V4). The output of the d-c restorer consists of highly negative pulses which occur at every other alternation of power-line frequency. These pulses are applied to the oscillator-doubler (V3) grid, and succeed in stopping the oscillator while the videosweep output of the instrument sweeps back from 10 mc to 100 kc. A baseline representing zero level is thus obtained on the oscilloscope trace during this period.

Power Supply — All plate and heater power requirements are supplied by a built-in power supply incorporating an electronic voltage regulator of conventional design. The instrument also supplies, at the SCOPE PHASING jacks, a 60-cycle voltage of adjustable phase for the horizontal deflection plates of an oscilloscope.

#### MAINTENANCE

2. Connect the WA-21A output cable, through a suitable detector, to the vertical terminals of the oscilloscope (same setup as Figure 2).

3. Set the SWEEP-MARKER-CW switch on "MARKER"; set the TUNING control for a marker frequency of 10 mc; place the BLANKING switch on "OFF."

4. Rotate the OUTPUT control to its approximate maximum position; turn the power switches of the oscilloscope and the WA-21A on.

5. A horizontal line should appear on the cathode-ray tube. This line may or may not contain the 10-mc marker and the zero-frequency beat, depending upon the degree of misalignment. If the zero-frequency beat does not appear on the line, then adjust the core of L17 (see Figure 13 for location) until the beat appears at the left-hand side of the trace.

6. Adjust the SCOPE PHASING control for a single, overlapping trace, then carefully reset the core of L17 until the zero-frequency beat rests near the left-hand side of the trace (this beat should not be set exactly on the extreme left-hand side of the trace, as any subsequent drift in the oscillators may cause the beat to disappear).

7. Throw the BLANKING switch to "ON."

8. Adjust the SWEEP WIDTH control on the rear apron until the 10-mc marker rests near the righthand side of the sweep. This adjustment may necessitate resetting the zero-frequency beat to the left-hand side of the trace by adjusting L17. These two adjustments (L17 and SWEEP WIDTH) must be made alternately until the zero-frequency beat rests near the left-hand side of the trace, and the 10-mc marker pip rests near the right-hand side (see Figure 12).

9. Adjust the core in L18 (see Figure 13 for location) until the trace is of maximum amplitude.

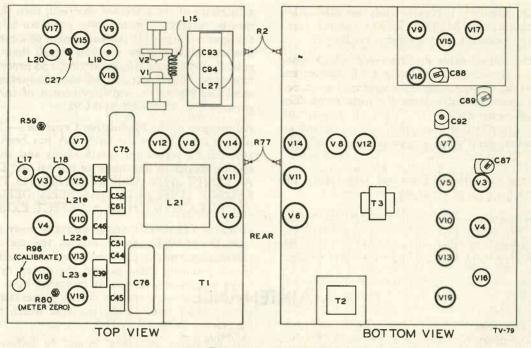


Figure 13—Tube and Adjustment Locations

Adjusting Bandwidth for Frequencies Less Than 10 Mc — If the Video Sweep Generator is to be used in testing equipment which has a bandwidth appreciably less than 10 mc, it may be desirable to adjust the instrument to sweep the required frequency band. By reducing the swept bandwidth, the trace of a narrow-band circuit can be made wider and thus more easily investigated. The general procedure for setting the WA-21A to sweep a desired frequency band can best be described by means of an example.

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Suppose it is desired to investigate the frequency characteristics of a 500-kc band-pass filter with a mid frequency of six megacycles. The WA-21A should be set to sweep a band of frequencies from about 5.5 to 6.5 megacycles. This is accomplished as follows:

1. Follow steps 1 to 4 described under the previous heading, "Sweep Output Bandwidth Adjustment."

2. Rotate the TUNING control until two marker pips appear on the trace, then adjust the SCOPE PHASE control until these pips coincide. Throw the BLANKING switch to "ON."

3. Retard the setting of the SWEEP WIDTH control until a band of frequencies one mc wide (6.5 mc-5.5 mc) is swept. This can be checked by setting the marker pip first on one side of the trace, then on the other, and noting the difference on the tuning dial scale.

4. Set the TUNING control for a frequency of 5.5 mc.

5. Adjust L17 (see Figure 13 for location) until the 5.5 mc marker pip rests at the left-hand side of the

trace. Adjust L18 (see Figure 13 for location) for a trace of maximum height.

6. The Video Sweep Generator is now set to sweep the frequency band from 5.5 mc to 6.5 mc, and the filter to be tested can be connected between the WA-21A output cable and the detector. Other frequency bands may be obtained in a manner similar to that described in the foregoing steps.

**Phasing the Blanking Circuit** — When the blanking circuit is correctly phased, the baseline of the oscilloscope trace will fall exactly under the upper portion of the pattern as shown in Figure 4. This trace is obtained on the oscilloscope as described under the heading, "Operation." An improperly phased blanking circuit is indicated when the reference baseline is displaced, as shown in Figure 14. To phase the blanking circuit, adjust R59 (see Figure 13 for location) for the pattern shown in Figure 4.

Voltmeter Calibration — The following procedure is recommended:

1. Turn the instrument on by rotating the SCOPE PHASING control clockwise.

2. Set the SWEEP-MARKER-CW switch to "CW." Set the TUNING control to about one mc.

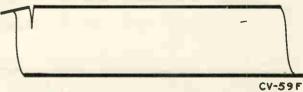


Figure 14—Improperly Phased Response

3. Turn the OUTPUT control fully counterclockwise, and adjust the METER ZERO control (see Figure 13 for location) for zero meter reading.

4. Insert the output cable into its receptacle on the rear apron, and connect an accurate a-c voltmeter to the leads on the output cable. This voltmeter must be flat at least up to one mc, and must read in rms value for sine-wave voltage.

5. Set the MULTIPLY BY switch to "2" and advance the OUTPUT control until the external voltmeter reads one volt.

6. Adjust the CALIBRATE control (see Figure 13 for location) for a panel-meter indication of 0.5.

7. Check the other positions of the MULTIPLY BY switch by comparing the panel-meter indication with the a-c voltmeter reading. The panel-meter reading times the MULTIPLY-BY switch setting should equal the a-c voltmeter reading for any setting of the OUT-PUT control or the MULTIPLY BY switch. If checks at lower levels are desired, direct comparison of the c-w output signal level with the output of a reliable signal generator may be made using a receiver with a signal-level meter. The input impedance of the receiver should be considered when calculating absolute signal levels at the cable ends of the WA-21A and the comparison signal generator.

Marker Oscillator Adjustment — The following procedure is recommended if the marker signal is suspected of being off frequency:

1. Set the SWEEP-MARKER-CW switch on "CW," turn the OUTPUT control fully clockwise, and apply power to the instrument by rotating the SCOPE PHASING control clockwise. Allow a half-hour warm-up period.

2. Couple a lead from a heterodyne frequency meter loosely to the wiring near the fixed marker oscillator tube socket (V9).

3. Tune the frequency meter to 22 mc and adjust L19 (see Figure 13 for location) for zero beat as noted on the frequency meter.

4. Remove the frequency-meter lead from the wiring near the socket of the fixed marker oscillator, and couple it to the output cable.

5. Tune the frequency meter to 9 mc, and set the TUNING control on the WA-21A to this same frequency. Adjust C27 (see Figure 13 for location) for zero beat (in the frequency-meter output).

6. Tune the frequency meter, and the TUNING control on the WA-21A, to one mc. Adjust L20 (see Figure 13 for location) for zero beat (in the frequencymeter output).

7. Repeat steps 5 and 6 until the tuning-dial scale on the WA-21A tracks properly.

**Spurious Responses** — Several trap circuits have been included in the instrument to eliminate spurious responses in the output of the instrument. If spurious responses, such as a stationary or a movable beat, or

a fuzziness of the trace are observed, then these traps may be carefully adjusted in an endeavor to eliminate the spurious signal. If the adjustment of a certain trap does not eliminate the spurious signal, then carefully reset the trap to its original setting, and proceed to the next one. The locations of these adjustments are shown in Figure 13, and they consist of trimmer capacitors C87, C88, C89, and C92.

Adjustment of the Modulation Capacitor — The modulation capacitor in the WA-21A has been factoryadjusted to very rigid specifications and should not require service in the course of normal use. DO NOT ATTEMPT ANY ADJUSTMENTS TO THIS AS-SEMBLY UNLESS IT HAS BEEN DEFINITELY ASCERTAINED THAT A DEFECT EXISTS.

If the modulation capacitor becomes noisy in operation, the trouble may be due to the coil assembly rubbing against the pole piece, or the capacitor plates shorting during some part of the sweep cycle. This latter condition will also be indicated by a break in the response curve of the Sweep Generator (or the circuit under test) at the point at which the short occurs.

The following procedure should be followed when the capacitor is serviced:

1. Remove the capacitor assembly from the WA-21A.

2. Remove the bakelite top plate by removing the four bolts ("A" in Figure 15). Place the top plate to one side; be careful not to damage the stator plates.

3. Apply 0.5 volt a-c to the unit. If the noise still continues, the coil is rubbing against the pole piece. (The noise which indicates faulty operation is different from the normal, smooth buzzing sound of a unit in good operating condition. It is louder and often erratic.) If the rubbing noise is no longer present when the stator assembly is removed, skip section 4 below and continue with section 5.

4. Loosen the four clamps, "C" in Figure 15, and, with the coil energized, shift the coil-support ring slightly until the noise disappears. Tighten the ring clamps, and make sure that the coil is not rubbing after the clamps are tightened. DO NOT LOOSEN ANY OF THE SUPPORT SCREWS ("N"), FOR THE STRAP CLAMPS OR THE CAPACITOR WILL PROBABLY BE DAMAGED BEYOND REPAIR.

5. Carefully replace the bakelite plate; do not tighten the mounting bolts.

6. The stator plates can be roughly centered by shifting the bakelite plate until the rubbing noise is not present. In order to insure maximum linearity, however, it will be necessary to center the stator plates exactly. The setup illustrated in Figure 16 is suggested for centering the stator.

7. Connect the vibrating capacitor as shown in Figure 16. Make the connecting leads to the capacitor plates as short as possible.

8. Apply 0.5 volt a-c to the terminals of the coil.

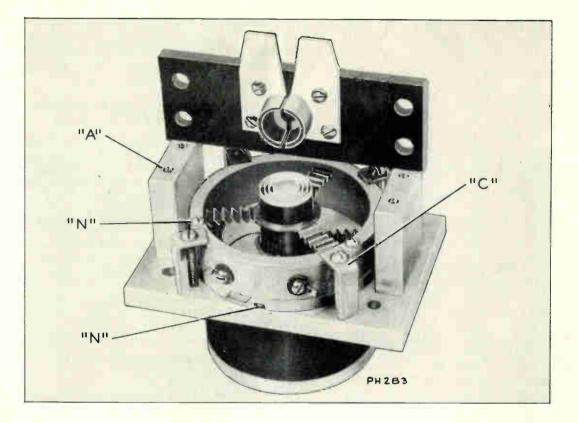


Figure 15-Modulation Capacitor

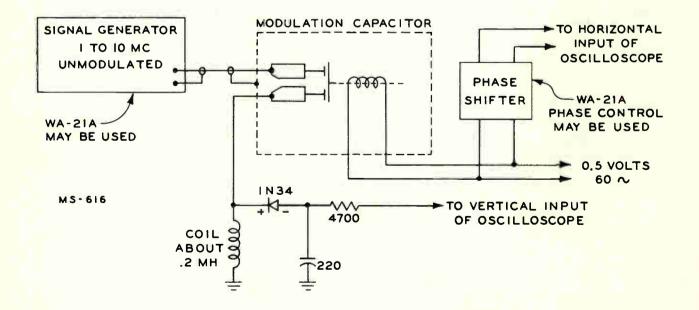


Figure 16—Modulation Capacitor Test Setup

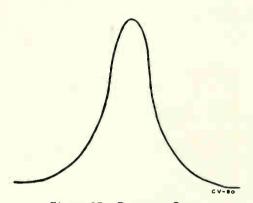


Figure 17—Response Curve

9. Adjust the signal generator output to approximately 0.1 volt.

10. Vary the frequency of the signal generator until the tuning of the vibrating capacitor and the test coil is found to sweep through the generator frequency (This should be approximately 4 mc for the coil shown in Figure 16.)

11. Adjust the phase shifter until the trace illustrated in Figure 17 is obtained.

12. Slowly increase the frequency of the signal generator and note in which direction the trace of Figure 17 moves.

13. Shift the bakelite plate until the curve is moved the maximum amount in the OPPOSITE direction

from that noted in step 12. At this point the stator plates will be most nearly centered within the movable plates.

14. Tighten the support bolts for the bakelite plate.

Voltage Measurements — Operating voltages should be within  $\pm 20\%$  of those indicated on the schematic diagram, Figure 18, when measured with a highresistance voltmeter and a power source of 117 volts a-c.

Resistance Measurements — All resistances should be within  $\pm 10\%$  of the values shown on the schematic diagram.

Voltage-Regulator Adjustment — Adjust the VOLT-AGE REGULATOR control (R77) on the rear apron for a d-c voltage of 260 volts, as measured from the cathode of either V11 or V14 to ground.

Fuse Replacement — Two cartridge fuses are located on the rear apron for protection against overload and internal short circuits. If fuse replacement is necessary, use only fuses with a 3-ampere rating.

Aligning the Video Amplifier — If, when the instrument is set up as described in the section on Operation, with the BLANKING switch set to "ON," the top line of the rectangular trace obtained is not straight and parallel to the bottom line (see Figure 4 for correct trace shape) then the video amplifier may need realignment. This is accomplished by adjusting L21, L22, and L23 (see Figure 13 for location) for the required flat-top response. This response should be obtained when the core studs are in their approximate mid-position.

## REPLACEMENT PARTS LIST

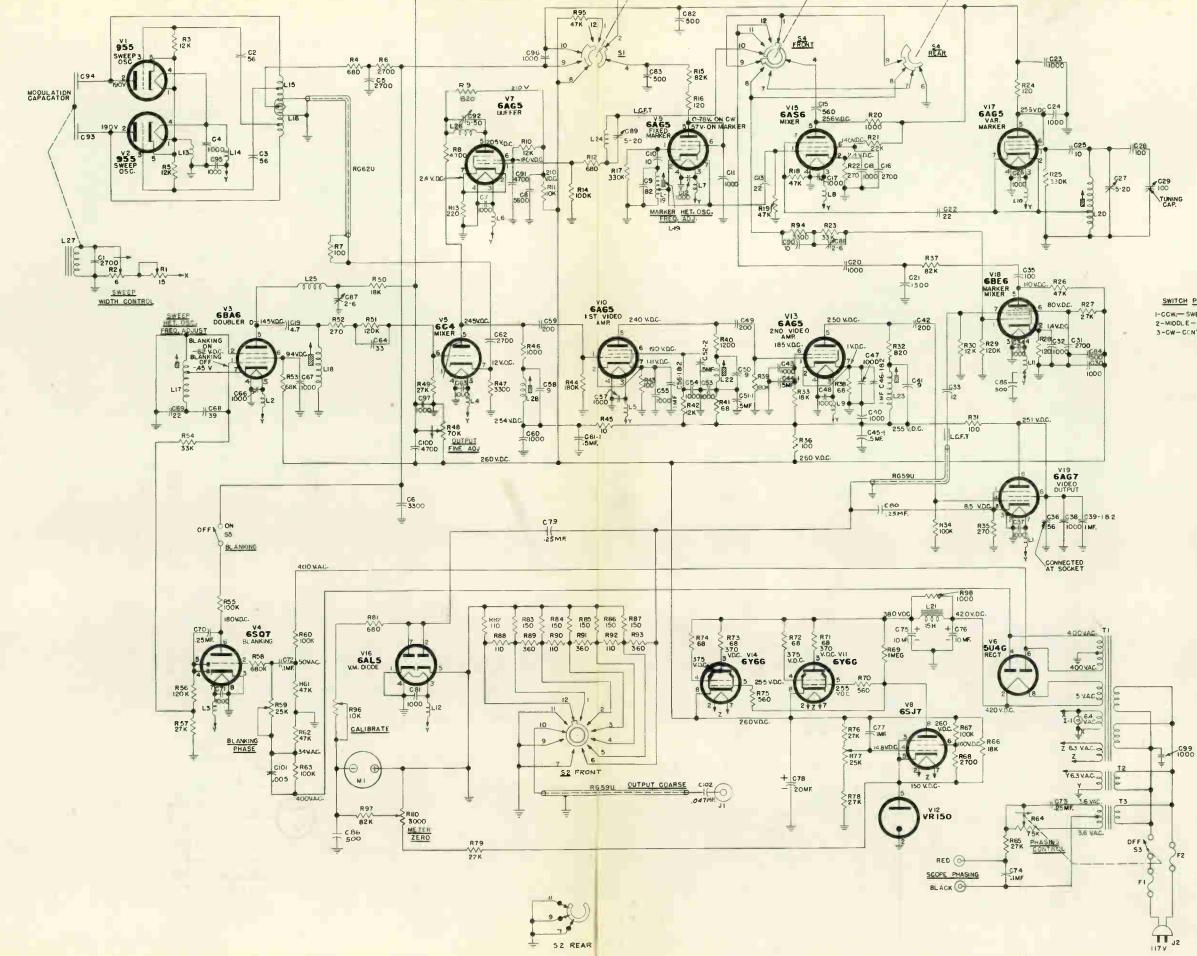
#### VIDEO SWEEP GENERATOR

#### TYPE WA-21A

Symbol No.	DESCRIPTION	Stock No.	Symbol No.	DESCRIPTION	Sto No
C-1	Capacitor-Mica, 2700 mmfd.,		C-64	Capacitor-Mica, 33 mmfd.,	
	±10% 500 Volts	39662		±10% 300 Volts	396
C-2, 3	Capacitor—Mica, 56 mmfd.,		C-66, 67	Same as C-4	
	±10% 500 Volts	39622	C-68	Capacitor—Mica, 39 mmfd.,	
2-4	Capacitor—Ceramic, 1000	71000	0.00	±10% 300 Volts	396
C-5	mmfd., ±20% 300 Volts	71929	C-69	Same as C-13	1
C-6	Same as C-1 Capacitor Mica 2200 mmfd		C-70	Capacitor—Tubular, .25 mfd.,	200
5-0	Capacitor—Mica, 3300 mmfd., ±20% 500 Volts	39664	C-71	$\pm 20\%$ 400 Volts	706
2-7	Same as C-4	39004	C-72	Same as C-4	
2-8	Capacitor-Mica, 5600 mmfd.,		0-12	Capacitor—Tubular, .1 mfd., ±20% 400 Volts	735
	±10% 500 Volts	39670	C-73	Same as C-70	133
2-9	Capacitor-Mica, 82 mmfd.,		C-74	Same as C-72	
	±10% 300 Volts	55328	C-75, 76	Capacitor-Paper, 10 mfd., 600	
2-10	Capacitor-Mica, 10 mmfd.,			Volts	527
	±10% 300 Volts	39604	C-77	Capacitor-Oil Treated, 1 mfd.,	
2-11, 12	Same as C-4			400 Volts	458
2-13	Capacitor-Mica, 22 mmfd.,	30612	C-78	Capacitor—Dry Electrolytic, 20	
2-15	$\pm 10\%$ 300 Volts Capacitor—Mica, 560 mmfd.,	39612	0.00	mfd., ±10% 450 V.	5624
	±10% 300 Volts	53111	C-79, 80	Same as C-70	
-16	Same as C-1	30	C-81 C-82, 83, 84,	Same as C-4 Conscient Food Thrue 500	
-17, 18	Same as C-4			Capacitor—Feed Thru, 500	
-19	Capacitor-Molded Head Lead		85, 86 C-87	mmfd. Capacitor Trimmer 2.6 mmfd	553
	Type, 4.7 mmfd., 500 V.	54402	C-87	Capacitor—Trimmer, 2-6 mmfd. Capacitor—Trimmer, 2-6 mmfd.	474
-20	Same as C-4		C-89	Capacitor—Trimmer, 5-20 mmfd.	
-21	Capacitor-Ceramic, 1500	50044	C-90	Same as C-10	010
	mmfd., ±20% 300 Volts	58244	C-91	Capacitor-Mica, 4700 mmfd.,	
-22	Same as C-13 Same as C-4			±5% 500 Volts	531
-23, 24 -25	Same as C-10		C-92	Capacitor—Trimmer, 5-50 mmfd.	546
-26	Same as C-4		C-93, 94	Part of L-27	
-27	Capacitor-Variable, Ceramic		C-95, 96, 97	Same as C-4	
	5-20 mmfd.	55301	C-99	Same as C-4	
-28	Capacitor-Mica, 100 mmfd.,		C-100	Same as C-91	
	±10% 300 Volts	39628	C-101	Capacitor—Paper Tubular, .005	706
-29	Capacitor-Variable, 7 to 100	10204	C 102	mfd., 1000 Volts D.C. Capacitor—Molded Paper, .047	700
	mmfd.	18384	C-102	mfd., ±20% 500 Volts	735
-30	Same as C-4 Same as C-1		J-1	Holder—Fuse Holder	485
-31 -32	Same as C-4		L-2, 3, 4, 5, 6,	Reactor-R.F. Choke Insulated,	
-33	Capacitor-Mica. 12 mmfd.,		7, 8, 9, 10,	1.8 MU. H. ±10%, 1.6 Ohms	
-55	±10% 500 Volts	39606	11, 12	±20% .56 Amp.	594
-34	Same as C-4		L-17	Transformer - Sweep Hetero-	
-35	Same as C-28			dyne Freq. Control Adjust-	
-36	Same as C-2			ing	555
-37, 38	Same as C-4		L-18	Transformer-Doubler	555
-39-1 & 2	Capacitor-Paper, Dual55	01577	L-19	Transformer-Frequency Con-	
40	mfd., 400 Volts	912//		trol Inductor	555
-40	Same as C-4 Capacitor—Ceramic, 9 mmfd.,		L-20	Transformer-Variable, Mark-	
-41	500 Volts	55571		er Osc., Freq. Control In-	555
-42	Capacitor-Mica, 200 mmfd.,		T 21	ductor	333
	±5% 500 Volts	39635	L-21	Coil—Choke, Impedance at 3 Volts, 60 Cy. A.C. and .175	
-43	Same as C-4			Amp. D.C. shall be 4050	
-44, 51	Same as C-39			Ohms	485
-45	Same as C-39		L-22, 23	Coil-Peaking	555
-46	Same as C-39		L-24	Coil-R.F. Choke	546
-47, 48	Same as C-4		L-25	Coil—R.F. Choke	555
-49	Same as C-42 Same as C-41		L-26	Same as L-24	
-50 -52, 61	Same as C-41 Same as C-39		L-27	Capacitor-Vibrator Capacitor	
-52, 61	Same as C-4			Assem. (Incl. C-93 & C-94)	555
-56	Same as C-39		L-28	Same as L-22	
-57	Same as C-4		R-1	Resistor—A djustable,	
-58	Same as C-41			Wire Wound 15 Ohms	485
-59	Same as C-42		R-2	Resistor—Variable, 6 Ohms,	EFF
-60	Same as C-4		Da	25 Watt	555
-62	Same as C-1		R-3	Resistor—Carbon, 12,000 Ohms $\pm 10\%$ <sup>1</sup> / <sub>2</sub> Watt	304
-63	Same as C-4				1.304

Symbol No.	DESCRIPTION	Stock No.	Symbol No.	DESCRIPTION	Stoc No
R-4	Resistor - Carbon, 680 Ohms		R-49	Resistor-Carbon, 27,000 Ohms,	
•••	$\pm 10\%$ 1 Watt	19233		±10% 1/2 Watt	3040
2-5	Same as R-3		R-50	Same as R-33	
			R-51	Same as R-29	
8-6	Resistor — Fixed Composition,	22955			
	2700 Ohms, ±10% 2 Watt	33855	R-52	Same as R-22	
R-7	Resistor — Carbon, 100 Ohms,		R-53	Resistor—Carbon, 68,000 Ohms,	
2-8	$\pm 10\% \frac{1}{2}$ Watt	34765		±10% 1/2 Watt	141
	Resistor - Carbon, 4700 Ohms,		R-54	Same as R-23	
	±10% 1/2 Watt	30494	R-55	Same as R-14	
2-9	Resistor — Carbon, 820 Ohms,		R-56	Same as R-29	
		30158	R-57	Same as R-49	_
	±10% 1/2 Watt	30130			
R-10	Same as R-3		R-58	Resistor — Carbon, 680,000	205
2-11	Resistor-Carbon, 10,000 Ohms,		7. 10	Ohms, ±10% 1/2 Watt	305
	±10% 1 Watt	71914	R-59	Resistor-Variable, Wire	
R-12	Resistor — Carbon, 680 Ohms,			Wound, 25,000 Ohms, Screw	
	±10% 1/2 Watt	12262		Driver Slot	555
2-13	Resistor — Carbon, 220 Ohms,		R-60	Resistor - Carbon, 100,000	
-13		5201	11-00	Ohms, ±10% 1 Watt	32
	$\pm 10\% \frac{1}{2}$ Watt	5201	D 61 62	Posistor Carbon 47 000 Ohma	52
R-14	Resistor - Carbon, 100,000	0070	<b>R-61</b> , 62	Resistor-Carbon, 47,000 Ohms,	
	Ohms, $\pm 10\% \frac{1}{2}$ Watt	3252	D co	$\pm 10\%$ 1 Watt	719
2-15	Resistor-Carbon, 82,000 Ohms,		R-63	Same as R-60	
	±10% 1/2 Watt	8064	R-64	Resistor - Variable Composi-	
-16	Resistor—Carbon, 120 Ohms,			tion, 75,000 Ohms, ±10%	
-10		30180		(Scope Phasing) (Includes	
	±10% 1/2 Watt	30189		S_2)	C.C.C
R-17	Resistor — Carbon, 330,000		Der	S-3)	555
	Ohms, $\pm 10\% \frac{1}{2}$ Watt	14983	R-65	Same as R-49	
-18, 19	Resistor-Carbon, 47,000 Ohms,		R-66	Resistor-Carbon, 18,000 Ohms,	
	±10% 1/2 Watt	30787		±10% 2 Watt	187
-20	Resistor-Carbon, 1,000 Ohms,		R-67	Same as R-60	
		34766	R-68	Resistor-Carbon, 2700 Ohms,	
21	±10% 1/2 Watt	34/00	11.00		144
-21	Resistor-Carbon, 22,000 Ohms,		P 60	$\pm 10\%$ 1 Watt	144
	$\pm 10\% \frac{1}{2}$ Watt	30492	R-69	Resistor-Carbon, 1 Megohm,	
-22	Resistor-Carbon, 270 Ohms,			$\pm 10\%$ 1 Watt	719
	±10% 1/2 Watt	30929	R-70	Resistor - Carbon, 560 Ohms,	
		00000		±10% 1 Watt	388
2-23	Resistor-Carbon, 33,000 Ohms,	20695	R-71, 72, 73,	Resistor - Carbon, 68 Ohms,	
	±10% 1/2 Watt	30685	74	±10% 1 Watt	369
R-24	Same as R-16		R-75		509
R-25	Same as R-17			Same as R-70	1.0
R-26	Same as R-18		R-76	Resistor-Carbon, 27,000 Ohms,	-
-27 .	Resistor — Fixed Composition,			$\pm 10\%$ 1 Watt	719
-21	Resistor - Fixed Composition,	44213	R-77	Same as R-59	
	27,000 Ohms, ±10% 2 Watt	44213	R-78	Same as R-76	
2-28	Same as R-16		R-79	Resistor-Carbon, 22,000 Ohms,	
2-29	Resistor - Carbon, 120,000				710
	Ohms, $\pm 10\% \frac{1}{2}$ Watt	30180	R-80	±10% 1 Watt	719
2-30	Same as R-3		10-00	Resistor - Variable, Composi-	
	Resistor-Carbon, 100 Ohms,		D at	tion, 3000 Ohms, $\pm 10\%$	716
2-31		31215	R-81	Same as R-12	
	±10% 1 Watt	31213	R-82	Resistor - Carbon, 110 Ohms,	
2-32	Resistor-Carbon, 820 Ohms,			±5% 1/2 Watt	379
	±5% 1/2 Watt	30158	R-83, 84, 85,	Resistor Carbon 150 Ohm	5/5
-33	Resistor-Carbon, 18,000 Ohms,		86, 87	Resistor - Carbon, 150 Ohms,	0.00
-33	±10% ½ Watt	3219		±5% 1/2 Watt	308
			R-88	Same as R-82	
-34	Same as R-14		R-89	Resistor — Carbon, 360 Ohms,	
2-35	Resistor-Carbon, 270 Ohms,	20407		±5% 1/2 Watt	320
	±10% 1 Watt	30497	R-90	Same as R-82	
2-36	Same as R-31		R-91		
	Same as R-15		R-92	Same as R-89	
2-37	Desisten Carbon 60 Ohma			Same as R-82	
-38	Resistor – Carbon, 68 Ohms,	34763	R-93	Same as R-89	
	±10% 1/2 Watt	34/03	R-94	Same as R-47	
-39	Resistor — Carbon, 180,000		R-95	Resistor-Carbon, 47,000 Ohms,	
	Ohms, $\pm 10\% \frac{1}{2}$ Watt	11959		±10% 1/2 Watt	307
40	Resistor-Carbon, 1200 Ohms,		R-96		501
-40	TCSISIOI - CHIDON, 1200 CHIMB,	30731		Resistor-Variable, W.W. 10,-	
	±5% 1/2 Watt		D 07	000 Ohms	563
-41	Same as R-38		R-97	Same as R-15	
2-42	Same as R-3		R-98	Resistor-Carbon, 1,000 Ohms,	1
2-43	Same as R-7			±10% 2 Watt	374
			S-1	Switch—Sweep Marker, 1 Sec-	
2-44	Same as R-39				
R-45	Resistor - Carbon, 10 Ohms,	24751		tion, 3 Position	555
	±10% 1/2 Watt	34761	S-2	Switch-Multiplier, 1 Section,	
2 46	Resistor-Carbon, 1000 Ohms,			7 Position	555
2-46	FO( I/ Watt	34766	S.A	Switch-Sweep Marker, 1 Sec-	
	±5% 1/2 Watt	01700	S-4		555
2-47	Resistor-Carbon, 3300 Ohms,		0.5	tion, 3 Position	333
	±10% 1/2 Watt	30733	S-5	Switch - Single Pole, Single	
40	Resistor-Variable Wire			Throw Turn Switch, 1 Amp,	
2-48	Wound (Output) 70,000			250 Volts (Blanking), 3	
			-	Amp., 125 Volts	300
	Ohms	55558		Amn 1/5 Voire	

Symbol No.	DESCRIPTION	Stock No.	Symbol No.	DESCRIPTION	Stoc No.
T-1	Transformer — Power — 50/60 Cycle	48556		Jack—Pin Jack (Blue) Jack—Pin Jack (Red)	5523 5523
<b>T</b> -2	Transformer-Filament - Out- line Pri., 117 V., 60 Cy., Sec.,			Knob — Control, For Sweep Marker, Tuning, Blanking	
T-3	6.3 Volts—6 Amps. Transformer—Pri., 115 Volts,	54726		Scope Phasing, Output Mul- tiply By. (Includes 2 Set	
	Sec., 6.3 Volts Sec. Amps. 1, Pri, VA 14 Adapter — Male Adapter for	64306		Screws) Lamp-Pilot Lamp 6-8 Volts	1726 1189
	Cable to Connector Cable—Test Cable with Clip	54246		Meter-RCA Styling, 0 to 200 Micro Amps. D.C., Calibrated	
	Connector Cable—Test Cable with Clip	54663		0 to .6 Volt Moulding — Shell for Output	5555
	Connector Clip—Alligator Clip	54664 53677		Cable consists of one Right and 1 Left Hand	4745
	Coil — Voice Coil, Suspension & Ring Assem.	5 <mark>5285</mark>		Plug—Output, Amphenol, Sin- gle Contact Chassis, Mount-	
	Connector — Motor Connector Base Connector — Single Contact	47594		ing Type Plug—Pin Jack	5180 4706
	(Male) For Output Cable Cord — Power, 96" Long-2	<mark>6634</mark> 4		Resistor — Fixed Composition, 75 Ohms, ±5% ½ Watt	3470
	Conductor — Right Angle, 2 Prong Male Plug—Standard			Socket—Pilot Light Assembly (Red)	485
	Straight Female 2 Contact Plug	52556		Socket—Tube, 8 Contact, Elas- tic Stop Nut Mounting Socket — Tube, Acorn Type,	6993
	Drive—Vernier Friction Drive (Tuning)	55559		Ceramic Base, 5 Contacts Socket—Tube, Miniature, 7 Pin	993
	Fuse-3 Amp. Cartridge Type Glass Body	10907		Type	5422



SWITCH POSITIONS I-CCW, SWEEP OUTPUT 2-MIDDLE-SWEEP PLUS MARKER 3-CW-CONTINUOUS WAVE CUTPUT

> SWITCHES SHOWN FULLY CLOCKWISE (IN CW POSITION). ALL RESISTANCE VALUES IN OHMS. ALL CAPACITANCE VALUES BELOW ONE IN MF. AND ABOVE ONE IN MMF. UNLESS OTHERWISE NOTED. ARROWS AT CONTROLS INDICATE CLOCKWISE ROTATION , ALL VOLTAGES MEASURED TO GROUND USING RCA VOLTOHMYST. LINE VOLTAGE - 117V. RMS .

> > Figure 18-Schematic Diagram

