

EVERYDAY ELECTRONICS and computer **PROJECTS**

MAY 1983

85p

FREE
TRANSISTORS



Apple II and BBC Micro REAL-TIME CLOCK

TEST GEAR 83
LAB AMPLIFIER

TRAIN
CONTROLLER

PLUS TWO DESIGNS
FOR YOUR FREE TRANSISTORS

Australia \$1.50 Malaysia \$4.95 IR £1.26 (inc V.A.T.)

Microboards
**SPECIAL
OFFER**

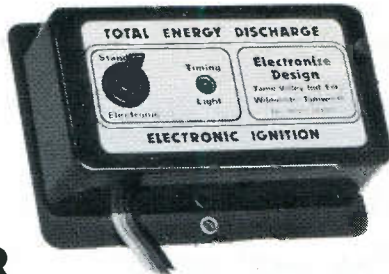


electronize

AUTO-ELECTRONIC PRODUCTS

KITS OR READY BUILT

ELECTRONIC IGNITION



IS YOUR CAR AS GOOD AS IT COULD BE ?

- ★ Is it **EASY TO START** in the cold and the damp? Total Energy Discharge will give the most powerful spark and maintain full output even with a near flat battery.
- ★ Is it **ECONOMICAL** or does it "go off" between services as the ignition performance deteriorates? Total Energy Discharge gives much more output and maintains it from service to service.
- ★ Has it **PEAK PERFORMANCE** or is it flat at high and low revs. where the ignition output is marginal? Total Energy Discharge gives a more powerful spark from idle to the engines max. (even with 8 cylinders).
- ★ Do the **PLUGS and POINTS** always need changing to bring the engine back to its best. Total Energy Discharge eliminates contact arcing and erosion by removing the heavy electrical load. The timing stays "spot on" and the contact condition doesn't affect the performance either. Larger plug gaps can be used, even wet or badly fouled plugs can be fired with this system.

★ Is the **PERFORMANCE SMOOTH**. The more powerful spark of Total Energy Discharge eliminates the 'near misfires' whilst an electronic filter smooths out the effects of contact bounce etc.

Most **NEW CARS** already have **ELECTRONIC IGNITION**. Update **YOUR CAR** with the most powerful system on the market - 3½ times more spark power than inductive systems - 3½ times the spark energy of ordinary capacitive systems, 3 times the spark duration.

Total Energy Discharge also features: **EASY FITTING, STANDARD/ELECTRONIC CHANGEOVER SWITCH, LED STATIC TIMING LIGHT, LOW RADIO INTERFERENCE, CORRECT SPARK POLARITY** and **DESIGNED IN RELIABILITY.**

★ **IN KIT FORM** it provides a top performance system at less than half the price of competing ready built units. The kit includes: pre-drilled fibreglass PCB, pre-wound and varnished ferrite transformer, high quality 2µF discharge capacitor, case, easy to follow instructions, solder and everything needed to build and fit to your car. All you need is a soldering iron and a few basic tools.

FITS ALL NEGATIVE EARTH VEHICLES
6 or 12 volt, with or without ballast.

OPERATES ALL VOLTAGE IMPULSE TACHOMETERS:
(Older current impulse types need an adaptor).

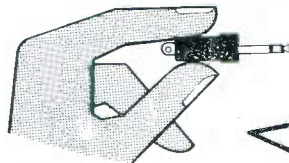
STANDARD CAR KIT £15-90
Assembled and Tested £26-70

PLUS
P. & P.
£1 (U.K.)

TWIN OUTPUT KIT £24-55
For Motor Cycles and Cars with twin ignition systems
Assembled and Tested £36-45

Prices
include
VAT

PROTECT YOUR CAR WITH AN ELECTRONIZE ELECTRONIC ALARM



Don't Wait Until
Its too Late ~
Fit one NOW!

- ★ **2000 COMBINATIONS** provided by an electronic key - a miniature jack plug containing components which must match each individual alarm system. (Not limited to a few hundred keys or a four bit code).
- ★ **60 SECOND ALARM PERIOD** flashes headlights and sounds horn, then resets ready to operate again if needed.
- ★ **10 SECOND ENTRY DELAY** allows owner to dis-arm the system, by inserting the key plug into a dashboard mounted socket, before the alarm sounds. (No holes in external bodywork, fiddly code systems or hidden switches). Re-closing the door will not cancel the alarm, before or after it sounds, the key plug must be used.
- ★ **INSTANT ALARM OPERATION** triggered by accessories or bonnet/boot opening.
- ★ **30 SECOND DELAY** when system is armed allows owner to lock doors etc.

★ **DISABLES IGNITION SYSTEM** when alarm is armed.

★ **IN KIT FORM** it provides a high level of protection at a really low cost. The kit includes everything needed, the case, fibreglass PCB, CMOS IC's, random selection resistors to set the combination, in fact everything down to the last nut and washer plus easy to follow instructions.

FITS ALL 12 VOLT NEGATIVE EARTH VEHICLES.
SUPPLIED COMPLETE WITH ALL NECESSARY LEADS AND CONNECTORS PLUS TWO KEY PLUGS

CAR ALARM KIT £24-95

ASSEMBLED AND TESTED £37-95

PLUS
P. & P.
£1 (U.K.)
Prices
include
VAT



Access and Visa
Welcome. Write or
Phone Quoting Number

Goods normally despatched
within 7 days

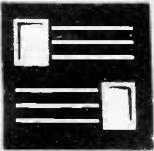
ELECTRONIZE DESIGN

Dept. C · Magnus Rd · Wilnecote
Tamworth · B77 5BY
tel: 0827 281000

EVERYDAY ELECTRONICS and computer PROJECTS

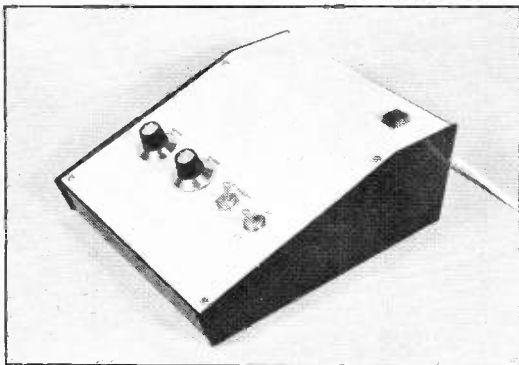
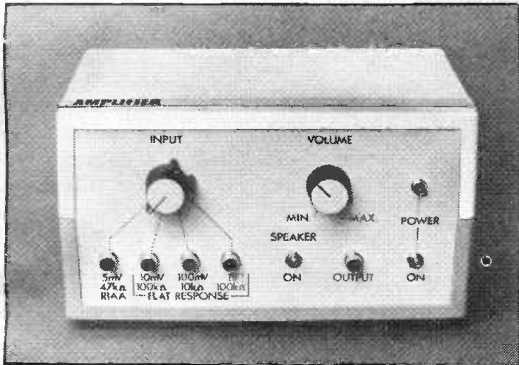
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FOR
YOUR
FREE
TRANSISTORS

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NO. 1 IN THE UK

The latest certificated ABC* figures show that Everyday Electronics is not only the largest selling electronic constructors' magazine in the United Kingdom, but has increased its lead by a very significant margin.

*Audit Bureau of Circulations.

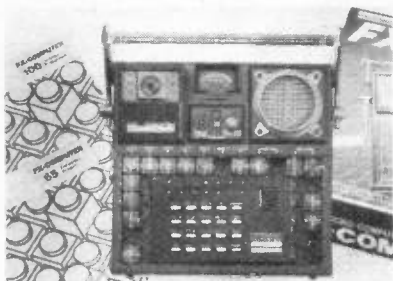
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Our June 1983 issue will be published on Friday, May 20. See page 303 for details.

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ELECTRONI-KIT

FX-COMPUTER



Teach-Yourself Computer and Electronics Construction Kit

A complete introduction to the "How, Why and What" of Computers and Electronics in the most practical way ever devised

THE KIT IS BATTERY-OPERATED AND COMPLETELY SELF-CONTAINED. NO TELEVISION OR OTHER EQUIPMENT IS REQUIRED. VERY EXTENSIVE MANUALS ARE INCLUDED

Ministry of Science and Technology, Japan — Prize Winning Product

The **FX-COMPUTER** is the ideal introduction to the study and understanding of computers and electronics. The kit offers remarkable versatility because the components are interchangeable and circuits are constructed by simply plugging specified components into the board provided in accordance with the instruction manuals. You quickly understand the principles involved and new circuits can be easily devised, built and dismantled. No soldering or wiring is involved, no tools are required; the components themselves complete the circuits.

No previous knowledge is required — very extensive educational manuals have been provided by English experts in computers and electronics. Working through the manuals you will soon be able to write programmes and "run" them and understand how computers work.

The following are just a few of the programmes in the Computer Manual (there are too many to list here) and also a few of the projects in the Electronics Manual:

How to Instruct the Computer and Store Information into Memories. Use of different instructions and Programming Techniques. Adding, subtracting, multiplying, dividing, averaging, counting up, counting down, etc. etc. — in Decimal and Hexadecimal. Converting Hexadecimal to Decimal, storing Random Numbers. Games: Tennis, Catch-the-Rat, Gun Fight, Slot Machine, etc. Using the Computer as a Musical Organ, storing and playing-back tunes, etc. **OVER 100 PROGRAMMES SHOWN IN THE COMPUTER MANUAL PLUS EXPLANATIONS AND DEMONSTRATIONS OF ALL TECHNICAL TERMINOLOGY.**

Electronic Components and How they Work — batteries, conductors, resistors, capacitors, diodes, transistors, lamps, photo-electric devices (CdS cell is included in the kit), oscillators, burglar alarms, control systems, organ, lie detector, etc., etc. **OVER 65 PRACTICAL WORKING PROJECTS SHOWN IN THE ELECTRONICS MANUAL**

All this is in ONE kit, costing about the same as the cheapest "Basic" ordinary Computer

The price is only £69.95 plus £3.00 P&P
(overseas rates quoted on request)

TRADE & EDUCATIONAL ENQUIRIES WELCOMED

Send cheque/PO/Access/Barclaycard to

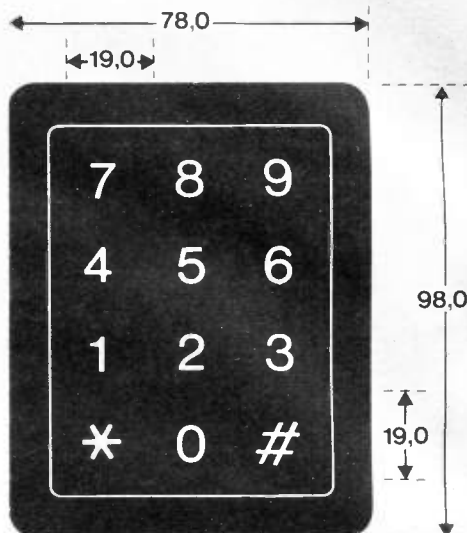
DEPT. EEFX. Electroni-Kit Ltd

It's not JUST a computer!

**ELECTRONI-KIT LTD.
388 ST. JOHN STREET
LONDON, EC1V 4NN (01-278 0109)**



3 × 4 KEY PAD



Features

- Sealed low profile construction
- Silver contacts on X-Y matrix
- For front or rear panel mounting
- Under surface printed graphics
- Easy clean and durable front surface
- Blue or red as standard options
- Available with P.C.B. or flexible backing

PLEASE ORDER BY PART NUMBER

Blue Flexible — FST 12\$0251	£6.50
Red Flexible — FST 12\$0252	£6.50
Blue P.C.B. — FST 12\$0351	£6.50
Red P.C.B. — FST 12\$0352	£6.50

Delivery by return — price includes V.A.T., postage & packing.

Send cheque or postal order to:-

CIRCUIT BOARD COMPONENTS

55 MURDOCK ROAD
BEDFORD
MK41 7PL

Telephone. (0234) 214219

GOING FOR A SONG!

GOING FOR A SONG — Are you a buyer for a factory, school, shop, club, etc? If so, please telephone or write for our sale list — over 1,000 tons of stock is being sold off at much below cost, simply because we must clear our big store this year. You will be amazed at the very low prices. Here are two examples:

LOT 1. is a parcel of approximately 1 million $\frac{1}{4}$, $\frac{1}{2}$ & 1 watt resistors. All uniformly packaged and front labelled in boxes ideal for retail display or self service racking. Normal stock valuation £6,000, offered at £950 the lot.

LOT 2. 1000 miniature encased relays — p.c.b. mounting — 2 change-over contacts 9 — 15V coil — very low current (700 ohm), normal stock valuation based on a similar relay R.S. price £2.56 each is £2,500. Parcel offered for £600.



8 POWERFUL MODEL MOTORS
(all different)
for robots, meccanos, drills, remote control planes, boats, etc. £2.95.

12 volt MOTOR BY SMITHS

Made for use in cars, etc., these are series wound and they become more powerful as load increases. Size $3\frac{1}{2}$ " long by 3" dia. They have a good length of $\frac{1}{4}$ " spindle — Price £3.45.
Ditto, but double ended £4.25.
Ditto, but permanent magnet £3.75.



EXTRA POWERFUL 12v MOTOR

Made to work battery lawnmower, this probably develops up to $\frac{1}{2}$ h.p., so it could be used to power a go-kart or to drive a compressor, etc. etc. £7.95 + £1.50 post.
(This is easily reversible with our reversing switch — Price £1.15).

SET OF 12v MOTORS!

Set of four 12 volt motors to drive passenger locomotive — £29.50
With address where to buy other parts.

WATERPROOF HEATING WIRE

60 ohms per yard, this is a heating element wound on a fibre glass coil and then covered with p.v.c. Dozens of uses — around water pipes, under grow boxes in gloves and socks. 23p a metre.

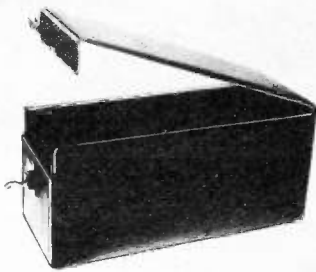
THERMOSTAT ASSORTMENT

10 different thermostats. 7 bi-metal types and 3 liquid types. There are the current stats which will open the switch to protect devices against overload, short circuits, etc., or when fitted say in front of the element of a blow heater, the heat would trip the stat if the blower fuses; appliance stats, one for high temperatures, others adjustable over a range of temperatures which could include 0 — 100°C. There is also a thermostatic pod which can be immersed, an oven stat, a calibrated boiler stat, finally an ice stat which, fitted to our waterproof heater element, up in the loft could protect your pipes from freezing. Separately, these thermostats could cost around £15.00 — however, you can have the parcel for £2.50.

POPULAR PROVEN PROJECTS

3 Channel sound to Light — with fully prepared metal case	£14.95
Ditto — made up	£25.00
Big Ear, listen through walls	£9.50
Robot controller — receiver/transmitter	£7.50
Ignition kit — helps starting, saves petrol, improves performance	£13.95
Silent sentinel Ultra Sonic Transmitter and receiver	£9.50
Car Light 'left on' alarm	£3.50
Secret switch — fools friends and enemies alike	£1.95
3 — 30v Variable Power Supply	£13.80
2 Short & Medium Wave Crystal Radio	£3.99
3v to 16v Mains Power Supply Kit	£1.95
Light Chaser — three modes	£17.50
Mullard Unilux Hi Fi stereo amplifier with speakers.	£16.75
Radio stethoscope — fault finding aid	£4.80
Mug Stop — emits piercing squawk	£2.50
Morse Trainer — complete with key	£2.99
Drill control kit	£3.95
Drill control kit made up	£6.95
Interrupted beam Kit	£2.50
Transmitter Surveillance Kit	£2.30
Radio Mike	£6.90
FM receiver kit — for surveillance or normal FM transmissions	£3.50
Seat Belt Reminder	£3.00
Car Starter Charger Kit	£14.00
Soil heater for plants or seeds	£16.50
Insulation Tester — electronic megger	£7.95
Battery Shaver or Fluorescent from 12v	£6.90
Matchbox Radio — receives Medium Wave	£2.95
Mixer Pre-amp — disco special with case	£16.00
Aerial Rotator — mains operated	£29.50
Aerial direction indicator	£5.50
40 watt amp — hi-fi — 20hz — 20KHz	£9.50
Microvolt multiplier — measure very low currents with ordinary multimeter	£3.95
Pure Sine Wave Generator	£5.75
Linear Power output meter	£11.50
115 Watt Amplifier 5hz — 25KHz	£13.50
Power Supply for one or two 115 watt amps	£17.50
Stereo Bass Booster, most items	£8.95

INSTRUMENT BOX



Instrument box with key. Very strongly made (plywood sides with hard board top and bottom). With black grained effect, vinyl covered, giving a very pleasing appearance. Internal dimensions 12 $\frac{1}{2}$ " long, 4 $\frac{1}{4}$ " wide, 6" deep. Ideal for carrying your multi range meter and small tools and for keeping them in a safe place. £2.30. Post paid if ordered with other goods, otherwise add £1.

LIGHTING & POWER CABLES

Copper clad, PVC sheathed. Made by Volex to BSS.

1.5mm single	per 100 metres	£2.30
1.5mm flat twin	per 100 metres	£4.50
1.5mm flat 3 core & E	per 100 metres	£5.50
4mm single	per 100 metres	£3.45
6mm flat 3 core	per 100 metres	£32.00
16mm flat twin & E	per 100 metres	£54.00

If not collecting, add 50% carriage.

SUB MINIATURE MOVING COIL SPEAKER OR MICROPHONE



Beautifully made permanent magnet type, this is only 0.8" dia. and 0.4" thick (approx). Impedance approximately 350 ohms, voice coil sealed with dust cover. Ex-equipment.
£1.15 each, or 10 for £10.

THIS MONTH'S SNIP

SUPER BARGAIN JUST ARRIVED: The AMSTRAD Stereo Tuner.

This ready assembled unit is the ideal tuner for a music centre or an amplifier, it can also be quickly made into a personal stereo radio — easy to carry about and which will give you superb reception.

Other uses are as a "get you to sleep radio", you could even take it with you to use in the lounge when the rest of the family want to view programmes in which you are not interested. You can listen to some music instead.

Some of the features are: long wave band 115 — 270 KHz, medium wave band 525 — 1650KHz, FM band 87 — 108MHz, mono, stereo switch, AFC switch, tuning meter to give you spot on stereo tuning, optional LED wave band indicator, fully assembled and fully aligned. Full wiring up data showing you how to connect to amplifier or headphones and details of suitable FM aerial (note ferrite rod aerial is included for medium and long wave bands. All made up on very compact board. This comes packed in a special container, which you can use as a temporary case if you want to make it quickly into a carry about personal stereo.

Offered at a fraction of its cost: **only £6.00**
+ £1.50 post + insurance.

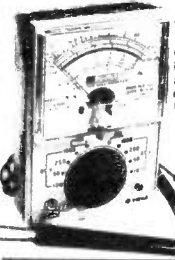
TANGENTIAL BLOW HEATER

2.5 Kw quiet, efficient instant heating from 230/240 volt mains. Kit consists of blower as illustrated, 2.5 Kw element, control switch and data all for £4.95. post £1.50.



MINI-MULTI TESTER Deluxe pocket size precision moving coil instrument, Jewelled bearings - 2000 o.p.v. mirrored scale. 11 instant range measures:

DC volts 10, 50, 250, 1000.
AC volts 10, 50, 250, 1000.
DC amps 0 — 100 mA.



Continuity and resistance 0 — 1 meg ohms in two ranges. Complete with test prods and instruction book showing how to measure capacity and inductance as well. Unbelievable value at only £6.75 + 60p post and insurance.

FREE Amps range kit to enable you to read DC current from 0 — 10 amps, directly on the 0 — 10 scale. It's free if you purchase quickly, but if you already own a Mini-Tester and would like one, send £2.50.

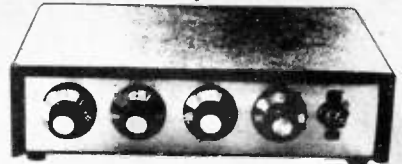
J. BULL (Electrical) Ltd.

(Dept. EE), 34 — 36 AMERICA LANE,
HAYWARDS HEATH, SUSSEX RH16 3QU. **Established 30 YEARS**

MAIL ORDER TERMS: Cash, P.O. or cheque with order. Orders under £10 add 60p service charge. Monthly account orders accepted from schools and public companies. Access & Barclay orders accepted day or night. Haywards Heath (0444) 454563. Bulk orders: write for quote. Delivery by return. Shop open 9.00 — 5.30, mon to Fri, not Saturday.

3 CHANNEL SOUND TO LIGHT

Now supplied with fully prepared metal work and p.c.b.
You'll have it going in an evening.



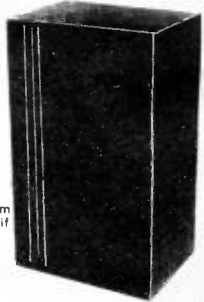
Complete kit of parts for a three channel sound to light unit controlling over 2000 watts of lighting. Use this at home if you wish but it is plenty rugged enough for disco work. The unit is housed in an attractive two-tone metal case and has controls for each channel, and a master dia. on/off. The audio input and output are by $\frac{1}{4}$ " sockets and three panel mounting fuse holders provide thyristor protection. A four pin plug and socket facilitate ease of connecting lamps. Special price is £14.95 in kit form or £25.00 assembled and tested. Case & metal Chassis No. Fully punched and prepared.

PANEL METERS & INSTRUMENTS

Singal meter, 200 VA	£1.15
Volt meter 0 — 200 volts, 2 $\frac{1}{2}$ " round	£1.75
Milli-amp meter. 500ma 2 $\frac{1}{2}$ " round	£1.15
— amp meter, hot wire scaled, 0 — 9 amps	£2.30
Diameter, 2 $\frac{1}{2}$ " round, centre zero, 500ma	£1.75
Charger panel meters, 1 $\frac{1}{2}$ " dia. sealed 3 amp	.75
Panel meter, 1 5/8" square, scaled VA	£1.15
Panel meter, Amstrad, 40mm sq. centre zero, scaled 1, 2, 3	.75
Edgeways panel, 3", 0 — 25ma, ex-GPO	£2.30

SUPER HI-FI SPEAKER CABINETS

Made for an expensive Hi-Fi outfit — will suit any decor. Resonance free. Cut-outs for 6 $\frac{1}{2}$ " woofer and 2 $\frac{1}{2}$ " tweeter. The front material is Dacron. The completed unit is most pleasing. Supplied in pairs, price £6.90 per pair (this is probably less than the original cost of one cabinet) carriage £3.00 the pair.



GOODMANS SPEAKERS

6 $\frac{1}{2}$ " 8 ohm 25 watt £4.50. 2 $\frac{1}{2}$ " 8 ohm tweeter. £2.50. No extra for postage if ordered with cabinets. Xover £1.50.

OTTO but for 8" speaker and 4" tweeter. £7.50 + £3.50.



VENNER TIME SWITCH

Mains operated with 20 amp switch, one on and one off per 24 hrs. repeats daily automatically correcting for the lengthening or shortening day. An expensive time switch but you can have it for only £2.95. These are without case but we can supply a plastic base £1.75 or metal case £2.95. Also available is adaptor kit to convert this into a normal 24 hr. time switch but with the added advantage of up to 12 on/off's per 24 hrs. This makes an ideal controller for the immersion heater. Price of adaptor kit is £2.30.

50 THINGS YOU CAN MAKE

Things you can make include Multi range meter, Low ohms tester, A.C. amps meter, Alarm clock, Soldering iron minder, Two way telephone, Memory jogger, Live line tester, Continuity checker, etc., and you will still have hundreds of parts for future projects. Our 10Kg parcel contains not less than 1,000 items — panel meters, timers, thermal trips, relays, switches, motors, drills, taps, and dies, tools, thermostats, coils, condensers, resistors, neons, earphone/microphones, nicad charger, power unit, multi-turn pots and data on the 50 projects.

YOURS FOR ONLY £11.50 plus £3.00 post.

EXTRACTOR FANS

Mains operated — ex-computer.

Woods extractor	4" x 4" Muffin 115v.
5" — £5.75, Post £1.25	£4.50, Post 75p.
6" — £5.95, Post £1.25	4" x 4" Muffin 230v.
5" Plannair extractor	£5.75, Post 75p.
£6.50, Post £1.25.	



ROTARY WAFER SWITCHES

5 amp silver plated contacts. $\frac{1}{4}$ " shaft, 1" dia. wafer.

Single wafer types, 29p each, as follows:		
1 pole 12 way	2 pole 6 way	3 pole 4 way
4 pole 3 way	6 pole 2 way	4 pole 3 way
Two wafer type, 59p each, as follows		
2 pole 12 way	4 pole 5 way	4 pole 6 way
6 pole 2 way	8 pole 3 way	12 pole 2 way
3 wafer types 99p each		
9 pole 4 way	6 pole 5 way	6 pole 6 way
	12p 3 way	18p 2 way

EXTRA POWERFUL 12v MOTOR

Made to work battery lawnmower, this probably develops up to $\frac{1}{2}$ h.p., so it could be used to power a go-kart or to drive a compressor, etc. etc. £6.90 + £1.50 post.
(This is easily reversible with our reversing switch — Price £1.15).

MINI MONO AMP on p.c.b. size 4" x 2"

approx. Fitted volume control and a hole for a tone control should you require it. The amplifier has three transistors and we estimate the output to be 3W rms. More technical data will be included with the amplifier. Brand new, perfect condition, offered at the very low price of £1.15 each, or 10 for £10.00.



Rapid Electronics

MAIL ORDERS:
Unit 3, Hill Farm Industrial Estate,
Boxted, Colchester, Essex CO4 5RD.
TELEPHONE ORDERS:
Colchester (0206) 36412.

ACCESS AND BARCLAYCARD WELCOME

LINEAR			
555CMOS 80	ICL7106 790	LM339 45	LM3911 120
555CMOS 150	ICL7611 95	LM348 60	LM3914 175
741 14	ICL8038 295	LM358 50	LM3915 195
748 35	ICL7621 180	LM381 65	MC1496 65
9400CJ 350	ICL8211A 200	LM3812 120	MC1496 65
AY-3-1270 720	ICM224 785	LM3832 120	MC1496 65
AY-3-8910 370	ICM7555 80	LM3904 130	ML922 400
CA3046 60	ICLF351 45	LM3904 130	ML924 195
CA3089 190	ICLF356 90	LM3931 100	ML926 140
CA3090AQ 375	ICL301A 360	LM3931 100	ML926 140
CA3130E 85	ICL301A 360	LM709 25	ML927 140
CA3140E 35	ICL301A 360	LM711 60	ML928 140
CA3161E 100	ICM324 40	LM725 350	ML929 40
CA3189 290	ICM3342 100	LM725 350	ML929 40
CA3240E 110	ICM3352 125	LM741 14	NE529 225
		LM741 14	NE531 225
		LM741 14	NE544 205
		LM741 14	NE555 110
		LM741 14	NE556 45
		LM741 14	NE556 110

TRANSISTORS			
AC125 35	BC149 9	BC547 7	BF800 23
AC126 25	BC157 8	BC549 10	BF800 23
AC127 25	BC158 8	BC549 10	BF800 23
AC128 20	BC159 8	BC549 10	BF800 23
AC176 25	BC172 8	BC549 10	BF800 23
AC187 22	BC186C 10	BC549 10	BF800 23
AC188 22	BC189C 10	BC549 10	BF800 23
AD142 120	BC170 8	BC549 10	BF800 23
AD149 80	BC171 10	BC549 10	BF800 23
AD161 40	BC172 8	BC549 10	BF800 23
AF124 60	BC178 18	BC549 10	BF800 23
AF126 50	BC179 18	BC549 10	BF800 23
AF139 40	BC182 10	BC549 10	BF800 23
AF186 70	BC182L 8	BC549 10	BF800 23
AF219 75	BC183 10	BC549 10	BF800 23
BC107 10	BC183L 10	BC549 10	BF800 23
BC107B 12	BC184 10	BC549 10	BF800 23
BC108 10	BC184L 7	BC549 10	BF800 23
BC108B 12	BC212 10	BC549 10	BF800 23
BC108C 12	BC212L 10	BC549 10	BF800 23
BC109 10	BC213 10	BC549 10	BF800 23
BC109C 12	BC213L 10	BC549 10	BF800 23
BC114 18	BC214 10	BC549 10	BF800 23
BC115 22	BC214L 8	BC549 10	BF800 23
BC117 18	BC237 8	BC549 10	BF800 23
BC119 25	BC238 14	BC549 10	BF800 23
BC137 40	BC308 12	BC549 10	BF800 23
BC139 40	BC327 14	BC549 10	BF800 23
BC140 28	BC328 14	BC549 10	BF800 23
BC141 30	BC337 14	BC549 10	BF800 23
BC142 25	BC342 25	BC549 10	BF800 23
BC143 25	BC477 30	BC549 10	BF800 23
BC147 8	BC478 30	BC549 10	BF800 23
BC148 8	BC479 30	BC549 10	BF800 23

CABLES			
20 metre pack single core connect-			
ing cable ten different colours, 65p			
Speaker cable 10p/m			
19 way grey ribbon 15p/m			
24mm twin screened 24p/m			
2.5A 3 core mains 23p/m			
10 way rainbow ribbon 65p/m			
10 way rainbow ribbon 120p/m			
19 way grey ribbon 80p/m			

HARDWARE			
PP3 battery clips 6			
Red or black crocodile clips 6			
Black pointer control knob 350			
PU Ultrasonic transducers 350			
12V Electronic buzzer 65			
12V Electronic buzzer 65			
PB3270 Piezo transducer 75			
64mm 54 ohm speaker 70			
64mm 8 ohm speaker 70			
20mm panel fusesolder 25			

CAPACITORS			
Polyester, radial leads, 250v, C280			
type 0.01, 0.015, 0.022, 0.033,			
6p, 0.047, 0.068, 0.1, 7p, 0.15,			
0.22, 9p, 0.33, 0.47, 13p, 0.68,			
70p, 1u - 23p			
Electrolytic, radial or axial leads:			
0.47/63V, 1/63V, 2.2/63V, 4.7/63V,			
10/25V, 9p, 22/25V, 47/25V, 8p,			
0.22, 9p, 0.33, 0.47, 13p, 0.68,			
220/25V, 50p			
Tag end power supply electrolytics:			
220/40V - 110p; 470/40V - 160p			
220/63V - 140p; 470/63V - 230p			
Polyester, miniature, 100V, 50p:			
1n, 2n, 3n, 4n, 6n, 8n, 10n, 15n, 7p,			
22n, 33n, 47n, 68n, 8p, 100n, 9p,			
150n, 11p, 220p, 33p, 330n, 20p,			
470n 26p, 680n, 29p, 1u 33p, 2u,			
50p.			

DIODES			
BY127 12	1N4001 3	1N4002 3	1N4003 3
0A47 10	1N4002 5	1N4003 5	1N4004 5
0A90 8	1N4006 7	1N4007 7	1N4008 7
0A200 8	1N5401 12	1N5402 12	1N5403 12
1N914 4	1N5406 17	1N5407 17	1N5408 17
1N4148 3	400MZen 6		

POTENTIOMETERS			
Rotary, Carbon track Log or Lin			
1K - 2M2, Single 32p, Stereo 85p,			
Single switched 80p, Slide 60mm			
travel 40mm Log or Lin 5K - 500K			
63p each.			
Preset submin. hor. 100 ohms - 1M			
7p each.			
Cermet precision multi-turn, 0.75W			
100 ohms to 100K - 85p each			

TRIACS			
400V 8A 65			
400V 16A 95			
400V 4A 50 BR100 25			

MULTIMETERS			
HT-120 4,000 opv			
A smart looking 11 range pocket			
sized multimeter with an impressive			
spec. Complete with battery, etc.			
650p each.			
HT-320 20,000 opv			
Highly sensitive 19 range multi-			
meter including transistor tester			
Overload protection. DC volts -			
1000, AC volts 1000, DC current			
0.25-4 resistance ranges. Complete			
with batteries, leads, etc. 1395p			

TRANSFORMERS			
Miniature plugs, 100V/2V all @ 100mA 100p each.			
PCB mounting. Miniature:			
3VA 0.6, 0.6 @ 0.25A, 0.9, 0.9 @ 0.15A; 0.12, 0.12 @ 0.12A 200p each.			
6VA 0.6, 0.6 @ 0.5A; 0.9, 0.9 @ 0.3A; 0.12, 0.12 @ 0.25A 270p each.			
High quality. Split bobbin construction:			
6VA 0.6, 0.6 @ 0.5A; 0.9, 0.9 @ 0.4A; 0.12, 0.12 @ 0.3A 220p each.			
12V 0.6, 0.6 @ 1A; 0.9, 0.9 @ 0.8A; 0.12, 0.12 @ 0.5A; 0.15, 0.15 @ 0.4A 295p (plus 40p carriage)			
25VA 0.6, 0.6 @ 1.5A; 0.9, 0.9 @ 1.2A; 0.12, 0.12 @ 1A; 0.15, 0.15 @ 0.8A 330p each (plus 60p carriage)			
0.8VA 0.12, 0.12 @ 2A, 0.15, 0.15 @ 1.5A. 440p each (plus 75p carriage)			

MIN. D CONNECTORS			
9 way 15 way 25 way 37 way			
Plugs solder lugs 60p 85p 125p 70p			
Right angle 120p 180p 240p 350p			
Sockets lugs 90p 130p 195p 290p			
Right angle 160p 210p 290p 440p			
Covers 100p 90p 100p 110p			

CONNECTORS			
DIN Plug Skt. Jack Plug Skt			
2 pin 9p, 5 pin 2.5mm 10p 9p			
3 pin 12p 10p 3.5mm 9p 9p			
5 pin 13p 11p Standard 16p 20p			
Phono 10p 12p Stereo 24p 25p			
12 pin 12p 4mm 14p 18p 17p			
UHF (CB) Connector 14p			
PL259 Plug 40p. Reducer 14p.			
SO239 square chassis skt 38p.			
SO239 round chassis skt 40p.			
IEC 3 pin 250V/6A.			
Plug chassis mounting 38p			
Socket free hanging 60p			
Socket with 2m lead 120p			

SWITCHES			
Submin toggle			
SPST 55p, SPDT 60p, DPDT 65p.			
Miniature toggle.			
SPDT 80p, SPDT centre off 90p,			
DPDT 90p, DPDT centre off 100p.			
Standard toggle:			
SPST 35p, DPDT 48p			
Miniature DPDT slide 12p.			
Push to make 14p.			
Push to break 22p.			
Rotary type adjustable stop.			
1P12W, 2P2W, 3P4W all 55p each.			
DIL switches:			
4SPST 80p 6 SPST 80p 8SPST			
100p.			

SOCKETS			
Low profile			
Wire-wrap			
8 pin 6p 25p			
14 pin 8p 35p			
16 pin 9p 42p			
18 pin 12p 52p			
20 pin 13p 60p			
22 pin 16p 70p			
24 pin 18p 70p			
28 pin 22p 80p			
40 pin 25p 98p			
Soldercon pins 60P/100			

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of 650 resistors)			
Ceramic Cap. kit. 5 of each value - 22p to 0.01u (135 caps)			
Polyester Cap. kit. 5 of each value from 0.01 to 1uF (65 caps)			
Resistor kit. Contains 5 of each value from 100 ohms to 1M (total			
of 425 resistors)			
50 68p resistors			
Nut and Bolt kit (total 300 items): 180p			
25 68A 1/4" bolts 50 68A washers 50 68A nuts			
25 68A 1/2" bolts 25 68A 1/4" bolts 50 68A washers			

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PACK 5 30 Low Profile IC Sockets 8, 14 and 16 — pin — 10 of each **£2.40**
PACK 6 25 Red LEDs (5mm dia.) **£1.25**

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AC176	23	BC348B	9	BY133	10	0 m m	330 7p; 2200 19p; 10V; 22 7p; 330
AC187	22	BC349A	10	OA47	8	2V7 to 47V	8p; 470 9p; 1000 16p; 2200 26p;
AC188	22	BC351	8	OA90	7	5p each.	16V; 10 4p; 22 5p; 47 6p; 100 7p;
AF124	55	BC394	19	OA91	7		220 8p; 470 12p; 1000 20p; 25V; 10
BC107	10	BC461	32	OA95	7	LEDs.	4p; 22 5p; 33 5p; 47 6p; 100 7p;
BC108C	12	BC477	28	1N4148	11	3mm Red	7 4p; 22 5p; 33 5p; 47 6p; 100 7p;
BC109C	12	BC478	29	1N4001	21	3 m m	220 10p; 330 12p; 470 14p; 1000
BC142	23	BC479	29	1N4002	31	Green	28p. 63V; 1 4p; 4.7 5p; 10 5p; 22
BC143	23	BC547	6	1N5401	9	3 m m	8p; 100 13p; 220 21p; 470 28p.
BC147A	8	BC547C	7	1N5402	10	Yellow	
BC148A	8	BC549	7	1N5404	13	3mm Clips 2	
BC149	8	BC549C	8				
BC168A	7	BC557B	8	POLYESTER CAPS C280 or Equiv.			
BC171A	8	BC558	8	0.01, 0.015, 0.022, 0.033 6p.			
BC171B	7	BC558B	8	0.047, 0.068, 0.1uF 7p, 0.15,			
BC177	18	BC570	16	0.22 9p. 0.33, 0.47 12p, 0.68,			
BC178	18	BC572	17	1uF, 1.5uF 20p.			
BC179	18	BD131	35	CERAMIC DISC CAPACITORS			
BC182	8	BD132	35	0.01uF 50V	2p	each	
BC182B	9	BD135	32	0.1uF 63V	6p	each	
BC182L	7	BD136	30	SEMICONDUCTOR MOUNTING KITS			
BC182LB	9	BD138	30	4 x TO3 Kits	20p	Pack	
BC183	8	BD139	35	4 x TO66 Kits	20p	Pack	
BC183L	8	BD140	35	4 x SO55 Kits	20p	Pack	
BC184	8	BD237	40	10 x TO126 Kits	20p	Pack	
BC184B	9	BD238	40	10 x TOP66 Kits	20p	Pack	
BC184L	7	BD243AX35	TIP2955	10 x TOP3 Kits	20p	Pack	
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BC213C	9	BF196	12	E12 Series 5%			
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BC214L	8	BF198	12	1/2 Watt 1Ω to 10MΩ	2p	each	
BC237B	7	BF199	12	100K Lin stereo Potentiometers 25p each.			
BC238B	9	BF246L	45				
BC258A	7	BF246LA	45				
BC206A	17	BF246LB	45				
BC261	18	BF259	30				
BC261A	18	BF337	35				
BC307	10	BF338	32				
BC307B	8	BF458	40				
BC308	10	BF459	40				
BC317	11	BF595	10				
BC327	12	BF595C	10				
BC328	12	BF595D	23				
BC337	11	BFY51	20				
BC338	12	BFY52	20				

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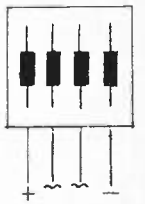
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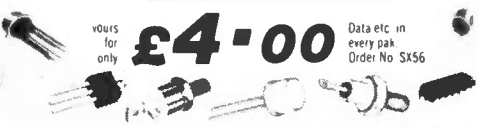
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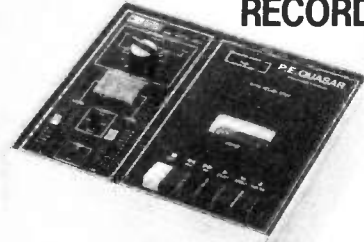
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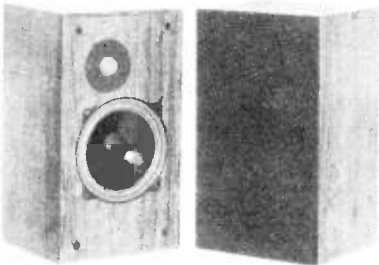
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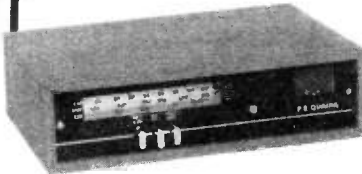
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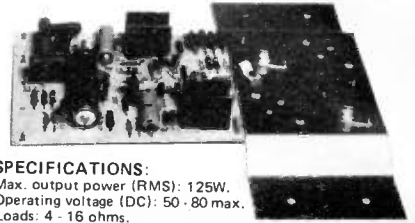
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VOL. 12 NO. 5 MAY 1983

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COMPUTER ADD-ON'S

THE development of electronic hardware add-on's for personal computers has opened up new areas and opportunities for the electronics constructor. Computer users with little or no experience of the practical work involved in our hobby may feel all this activity beyond them. They should not balk at the task: the rewards will greatly outweigh the time and effort spent in acquiring the necessary skills and expertise to build projects. Our constructional articles make everything crystal clear and back-up information concerning sources of supply for unusual or "difficult" components is given in *Shop Talk* every month. Features for the newcomer to electronics form an on-going part of our contents.

Add-on's and other units tailored to suit individual machines, or groups of machines, are already written into our publishing program. These designs are appearing on a regular basis now, and by judiciously ringing the changes we shall endeavour to keep all personal computer users happy, whether they be Apple, BBC, Genie, Pet or ZX aficionados. For no matter what machine is specifically catered for, many of these designs will be capable of wider application; furthermore, they will keep all computer users up to date with regard to the attractive propositions developing in the hardware region. Remember one good idea usually gives rise to many others . . . so if you have a personal computer you owe it to yourself to watch our pages regularly.

Electronic hardware is our business, however in the case of many computer add-on's it is not possible to escape the need for software. Providing that the essential listings are not of inordinate length they will be published in the text. Larger listings will be available directly from EE on cassette. Details of our newly opened Software Service appear elsewhere within this issue.

Computers are important. But they are only part of the electronic scene, as is borne out by this month's project repertoire. Apart from our Apple, BBC and Vic20, Pet projects we also cover such diverse interests as pop music (Guitar Headphone Amplifier) and model railways (Train Controller); then there's a small Personal Radio, a Moisture Detector and number three in our Test Gear 83 series. What is more, we go even further than just providing design information, for attached to the front cover is a pair of transistors which can be used in either of two particular projects.

Finally, just a hint about next month. For those who use a TRS-80, a Video Genie or a Caravan—there's something of special interest in our June issue. For further details please refer to page 303.



Readers' Enquiries

We cannot undertake to answer readers' letters requesting modifications, designs or information on commercial equipment or subjects not published by us. All letters requiring a personal reply should be accompanied by a stamped self-addressed envelope.

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Readers should note that we do not supply electronic components for building the projects featured in EVERYDAY ELECTRONICS, but these requirements can be met by our advertisers.

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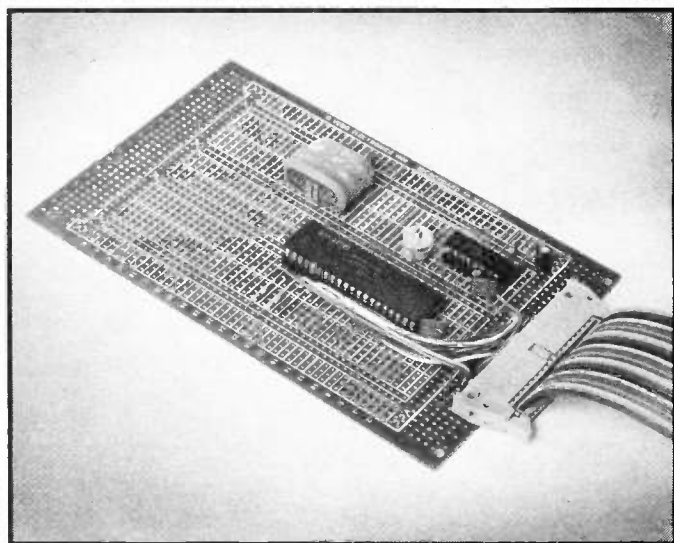
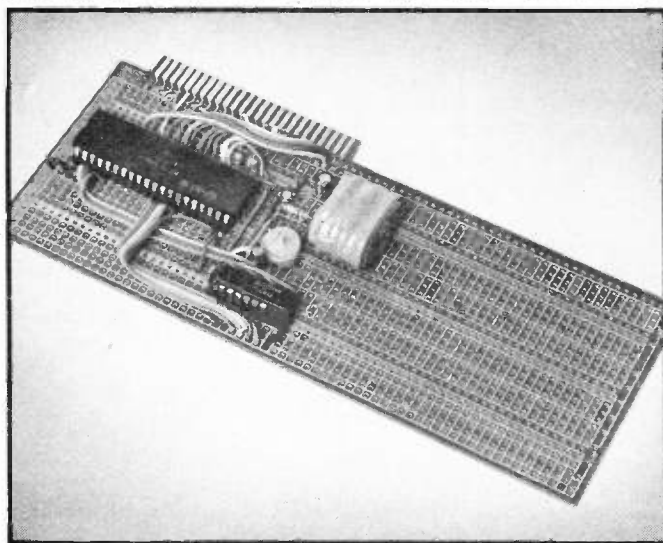
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REAL-TIME CLOCK APPLE II & BBC MICRO



Useful hardware add-ons for two popular micros. Capable of displaying day and time of day (hours, minutes and seconds).

BY O.N. BISHOP

WHEN you first hear someone refer to "real time" you may wonder why they bother to say "real". There is no doubt that time is real. Why the emphasis on its reality?

Real-time is a phrase used in the world of computers to distinguish between the actual time of day, as can be shown on the dial of any ordinary clock, and the various other kinds of time that mainly concern the computer itself. Ordinary clocks tell us the real time (half-past three, for example). The other clocks (or oscillators) in the computer are used for

telling it which stage it has reached in its cycle of operations.

Egg-timers and oven-timers are clocks which can be classed with the other clocks of the computer. They tell you what stage has been reached in the cycle of operations involved in boiling an egg or cooking a joint. Usually they tell you when the cycle is complete and the egg or joint is ready to eat.

It is not easy to use this kind of clock to tell you the real time, such as when you should switch on the TV to view your favourite programme.

SOFTWARE CLOCKS

All computers, even the simplest of micros, need a clock to time the operation of the system. Usually this is a high-frequency crystal oscillator, wired to the microprocessor. Few microcomputers have real-time clocks, although some such as the BBC Microcomputer allow you to make use of the system clock to control a special "clock program" which displays real time. This is a *software* clock, for it is not a physical part of the computer and must be loaded into

memory from a cassette tape (or keyboard) on every occasion that you use it.

HARDWARE CLOCK

The Real-Time Clock described here is a *hardware* clock, consisting of a special clock i.c. with another i.c. to interface it to the computer. The clock i.c. has many of the characteristics of the i.c. in a digital watch. The main difference is that the i.c. in a watch is designed to drive a 7-segment display. The outputs required for this are very different from those which a microprocessor requires. A microprocessor could be programmed to read the output intended for a 7-segment display, but the program required to do this would be unnecessarily complicated.

This project is designed specially for the Apple II personal computer and the BBC Micro. Owners of the Apple are fortunate in having a set of "slots" (sockets) on the circuit board of the micro into which they can plug circuit boards (or "cards") carrying interface circuits of various kinds.

Having made up the circuit as described, you simply open the case of the Apple, plug in the card, and the real-time clock is ready for use.

The socket which takes the card has been provided with connections to the data bus, the address bus and several

control lines. By using these we are able to make use of the on-board address-decoding facilities of the Apple, and so simplify circuit construction a great deal.

For the BBC Micro, a socket is required to be fitted to the circuit board and a cable to connect to the BBC 1MHz Bus (PL11).

Readers will be able to adapt this design to suit other micros. The main difference is that you will have to provide an address-decoding circuit. This is already provided by the Apple and the BBC Micro.

REAL-TIME CLOCK I.C.

The main sections of the clock i.c., an MM58174, are shown in Fig. 1. The crystal oscillator runs at 32.768kHz. Its frequency is first divided by 16/15 to give 30.720kHz and is then divided 9 times by a series of binary counters to give 60Hz. This is further divided by 6 to give the fundamental timing frequency of 10Hz, or tenths of a second.

The fundamental frequency is then fed to counters which register tenths, units and tens of seconds, units and tens of minutes, units and tens of hours, units and tens of day, day of the week, day of the month, and units and tens of years. The "year" register holds a value which depends on the position of the current year in the leap-year cycle.

Any of these registers (except the three "seconds" registers) may be written into to set the time of the clock when we first begin to run it. Any register (except the year register) can be read from when we want to know the time.

Register 14 is used to stop and start the clock. Register 15 is for interrupt status. This is for using the clock as an elapsed-time clock. It can be set to interrupt the micro once after a given period (which may be 0.5s, 5s or 60s), or repeatedly every 0.5s, 5s or 60s, as explained later.

Register 1 is used when testing the i.c. in the factory and must always be set to "0" before the clock is started, and every time the computer is switched on.

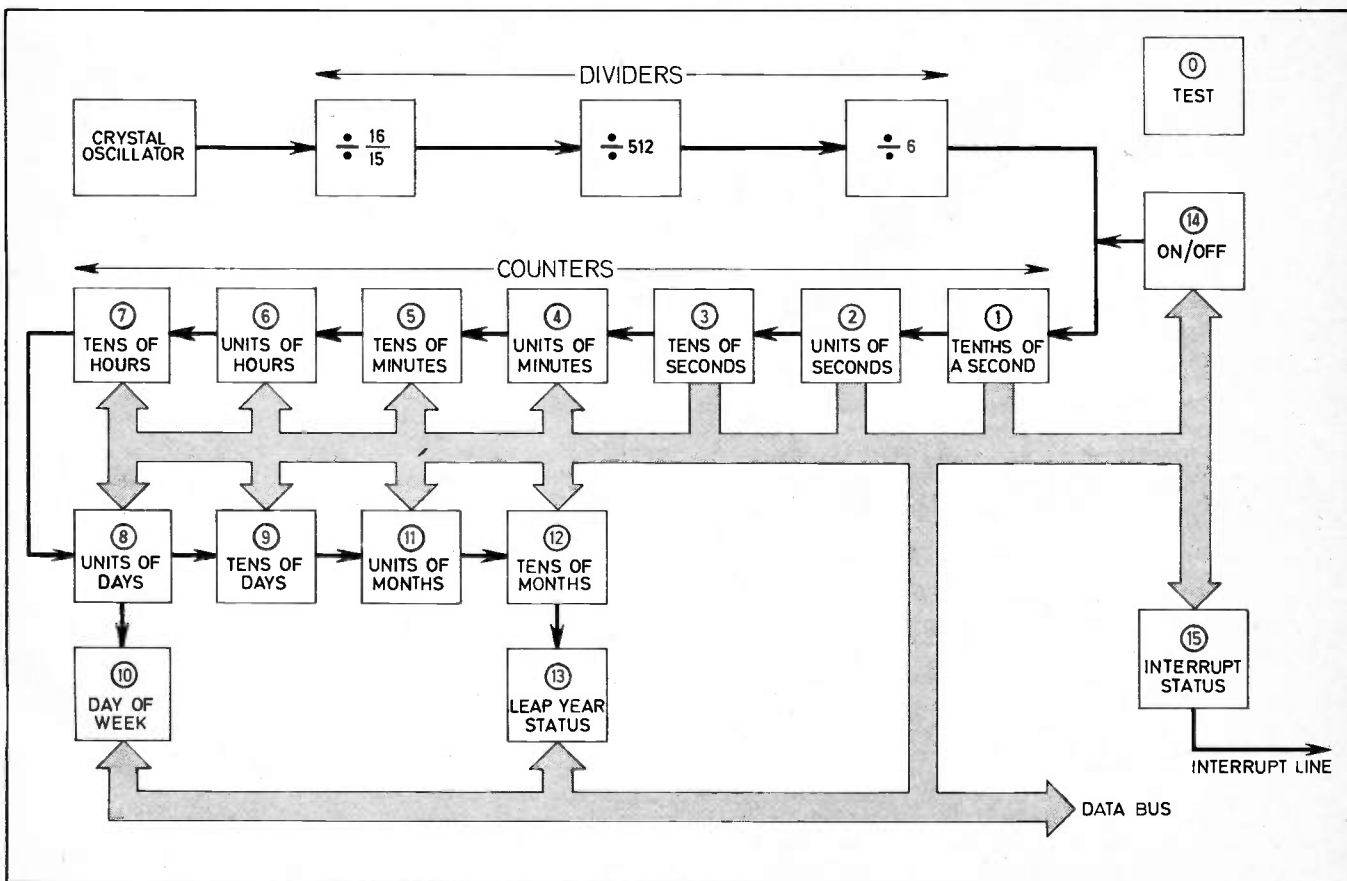
CIRCUIT DESCRIPTION

The circuit diagram is shown in Fig. 2. This has been drawn for the BBC Micro, with the different pinning connections to suit the Apple.

The circuit shows that the clock i.c. has a back-up battery. This is a big advantage, for the clock continues to run while the micro is switched off and if the board is unplugged from the computer.

Unlike a software clock, it does not need resetting every time the micro is turned on. When the micro is running and the i.c. is being powered by the +5V supply of the micro, it takes about 4mA, but

Fig. 1. Block diagram of the MM58174 Real-Time Clock i.c. The encircled numbers are the address within the clock of each of its registers.



when the micro is switched off and the clock is running from the battery, it takes only 4µA. The recommended nickel-cadmium battery has a capacity of 10mAh, so the clock will run for many months on the battery alone. When the micro is on, the battery is trickle-charged with a current of about 1mA.

PERIPHERAL INTERFACE ADAPTOR

In order to be able to use the clock i.c., we must look into the details of how the MC6821 (IC1) Peripheral Interface Adaptor (or PIA) works. One of the problems of using the clock i.c. is that the exact times at which data may be exchanged between the clock i.c. and the microprocessor inside the computer are very limited in extent and are determined by the cycle of operation of the micro.

It makes programming less complicated if we use a PIA. Data from the MPU which is intended for the clock i.c. may be fed to the PIA, which holds it ready for when the clock is able to read it.

The PIA is essential in transfer of data in the other direction, too. The clock can be made to send data to the PIA, but at that stage it is not placed on the data bus. To put data on to the bus at the wrong time would certainly upset the operation of the computer. Once in the PIA, the data

can be read by the MPU at a time convenient to itself.

It is apparent from the description above that the PIA is capable of sending data in either direction: micro-to-clock (output) or clock-to-micro (input).

TWO EIGHT-BIT PORTS

The PIA has a set of eight terminals to connect it to the MPU through the data bus. It has two sets of lines (called Port A and Port B), for communicating with the clock. Each port has eight lines, though we use only the lower four lines of each port (PA0 to PA3 and PB0 to PB3) in this circuit. The lines can pass data in either direction, but we must first decide upon which direction. This is done by setting the Data Direction Registers in IC1 for each port (DDRA and DDRB). The DDRs have eight locations (or bits), one corresponding to each line of the port. If a bit is set to "0", the corresponding line is defined as an *input*; if it is set to "1" it becomes an *output*.

In this circuit we define the four lower lines of Port A as outputs. These are used to output an address (in the range 0 to 15) to IC2, the clock i.c., this address being a number in the range 0 to 15, to select which register is to be read from or written to.

The four lines of Port B are used for sending data in either direction. They are defined as outputs when we are setting the clock and as inputs when we want to know the time.

CONTROL REGISTERS

The operation of each port is controlled by a control register (CRA, CRB). One of the uses of this register is to put us into communication with DDRA and DDRB, or with the registers which are used in transferring data between the micro and the clock. Table 1 shows the settings.

Table 1 also shows how the control registers are used to operate the control outputs CA2 and CB2. These lines are normally kept at high level (+5V), but CA2 is made low when we want to make the RD input of the clock low so as to be able to read data from it. CB2 is made low to make the WR input low when we want to write to the clock.

The RD input is held high by pull-up resistor R3 wired to the back-up battery. This holds RD high while the power supply is off. A low RD halts the counting process, so if R3 was not present, the clock time would not advance while power is off. The WR input has an internal pull-up resistor.

Fig. 2. Lower right shows the circuit diagram for the Real-Time Clock with connection details to the 1MHz bus of the BBC Micro; lower left shows the connections between circuit and the Apple II (slot 3).

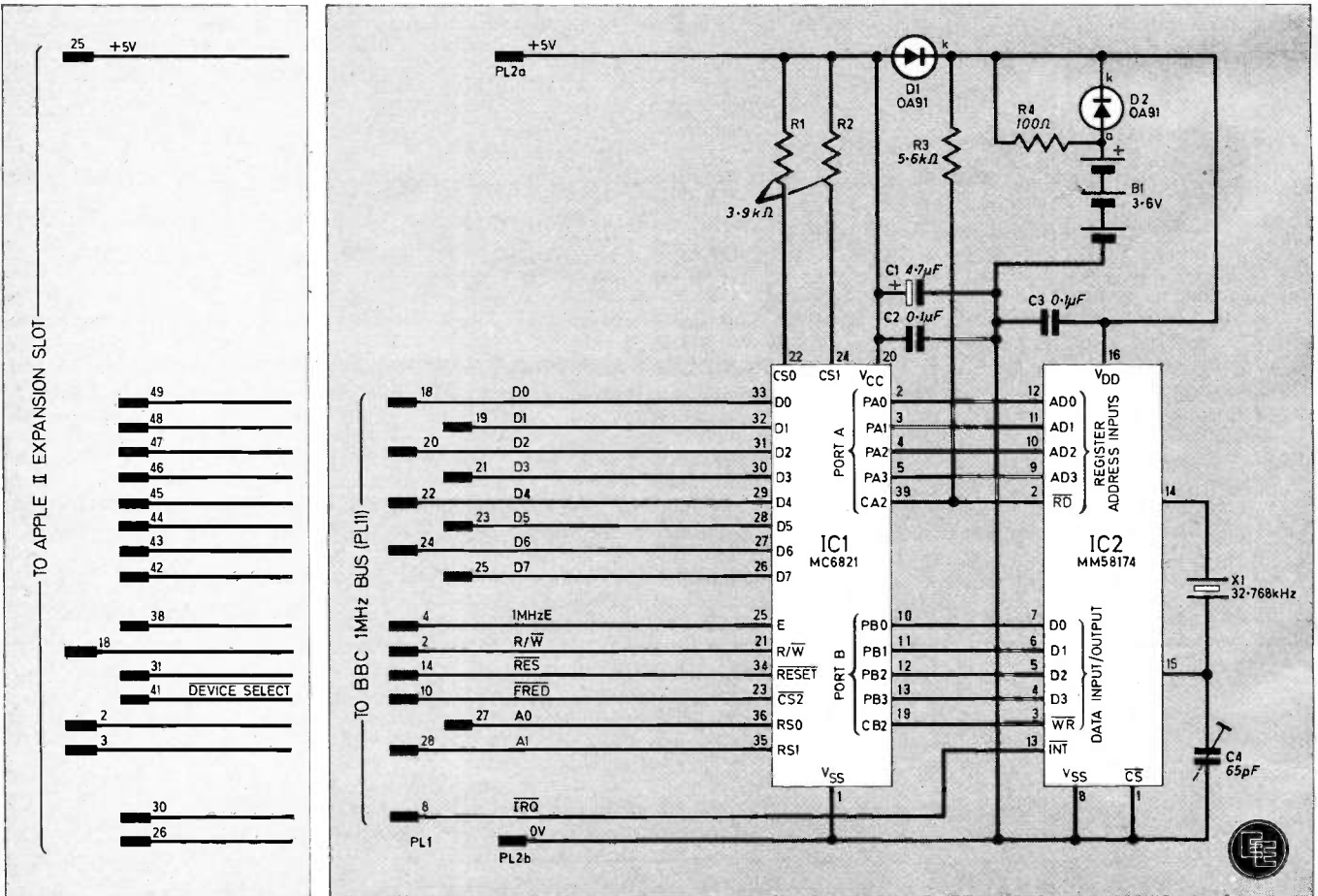


Table 1. Using the Control Registers of the PIA

Bit	7	6	5	4	3	2	1	0
Function	Not applicable		Controls output CA2			DDRA	Not applicable	
CA2 high with access to DDRA	0	0	1	1	1	0	0	0
	(Equivalent to 56 in decimal)							
CA2 high with access to Port A	0	0	1	1	1	1	0	0
	(Equivalent to 60 in decimal)							
CA2 low with access to Port A	0	0	1	1	0	1	0	0
	(Equivalent to 52 in decimal)							

The table refers to CRA, but the same coding applies to CRB.

DEVICE SELECT

In order to alter any of the registers in the PIA, the chip must be selected. In the Apple, this is done by using the **DEVICE SELECT** line of its slot. This line goes low whenever any address in the Peripheral I/O space of the slot is addressed. The programs given here assume that the clock card is in slot 3.

The Peripheral I/O space for this slot consists of the sixteen addresses \$C0A0 to \$C0AF (the "\$" indicates a hexadecimal number). **DEVICE SELECT** is wired to a "Chip Select" input (CS2) of the PIA. We need only four addresses (\$C0A0 to \$C0A3) to address the PIA, as shown in Table 2. The lower two address lines (A0 and A1) are wired to the "Register Select" inputs, RS0 and RS1. When one of the addresses is on the bus, CS is made low and the levels at RS0 and RS1 determine which register is put in communication with the data bus.

The other connections to the PIA are: CS0 and CS1: alternative "Chip Select" inputs, which require a high input to make the chip active, and are held high permanently by R1 and R2.

E: the Enable input is connected to the ϕ 1 clock line in the Apple; R/W: the read/write line is connected to the R/W line of the Apple; RESET: a low level on this line (as when the micro is first switched on or the Reset key is pressed) clears all registers of the PIA to zero (but not those of IC2).

In the Apple, the clock uses the +5V regulated supply which is provided at the slot. Your micro may be able to provide the current (about 100mA) from its own supply. You should check in the manual of your computer to ascertain how much current it can supply without overloading. If it cannot supply enough you will need to add a +5V regulator circuit to your clock board.

THE BBC MICRO

The BBC Microcomputer has the same MPU as Apple II (a 6502) which provides the same signals for controlling the clock. Table 3 shows which pins of the 1MHz bus connector to use.

The 1MHz bus is available on both Model A and Model B micros, so the clock circuitry can be used with either model.

Pin 4 (1MHzE) gives a clocking pulse which is used instead of ϕ 1 of the Apple. Pin 10 gives a signal called **FRED**, which is the equivalent of the **DEVICE SELECT** used on the Apple.

The power supply is taken from the power outlet on the underside of the BBC Micro, PL10.

The **FRED** output is normally high, but goes low when any address in the range &FC00 to &FCFF is used ("&" represents a hexadecimal number follows). Thus without further address decoding we can use &FC00 as the base

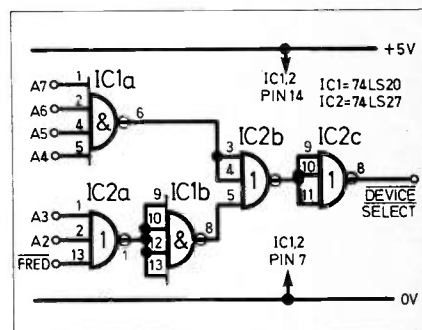


Fig. 3. Decoder circuit for use with the BBC Micro to place the Real-Time Clock at address &FCF0.

address of the clock (equivalent to address 3000 of Table 2, or 16205 in the Apple listings).

FURTHER DECODING

The snag about this system is that you will not be able to attach anything else to the bus at the same time. If you want to have more than one device attached, you must provide further address decoding. The built-in decoder of the BBC Micro takes care of the upper eight address lines, and the lower two go directly to IC1, so only the six lines A2 to A7 need decoding. Fig. 3 shows a circuit which gives the base address &FCF0.

Table 2. Addresses used by the PIA (assuming the base address is 3000)

Address	Register accessed
3000	Port A input/output OR* DDRA
3001	Control Register A (CRA)
3002	Port B input/output OR* DDRB
3003	Control Register B (CRB)

*Depends on the setting of bit 2 in the Control Register for the Port (see Table 1).

Table 3. Connections to the 1MHz Bus of the BBC Microcomputer

Pin No.	Name of line
2	R/W
4	1MHzE
10	FRED
14	RESET
18-25	D0-D7 (data bus)
27-34	A0-A7 (address bus)

APPLE II REAL-TIME CLOCK CONSTRUCTION



APPLE MICROBOARD

For use with the Apple II personal computer the circuitry is built on an Apple Compatible Microboard. This has been specially designed to plug directly into the extension "slots" on-board the Apple. It has a gold-plated finger set, and this design is intended to plug directly into slot 3. The board has a unique feature of a Colander Ground Plane for maximum screening purposes, and has solder mask protection.

The layout of the components is not critical and may be changed to suit requirements. Our layout design has been accommodated on one half of the board to allow other user-designed circuitry to be added at any time.

The layout of the components on the topside of the board is shown in Fig. 4, including link wires. There are no breaks to be made on the underside. Note that there is a through-the-board link to be made at the bottom right-hand corner of the board as drawn.

Two underside link wires are to be connected. There is one to join D1(k) to D2(k), and a second adjacent to this to connect IC1 pin 39 to IC2 pin 2.

Begin construction by soldering in both i.c. sockets followed by the shorter link wires and the remainder of the components in any order. Pay attention to the polarities of B1, D1, D2 and C1. The "leads" on B1 and C4 will need to be "thinned" using a small file so that they comfortably fit the hole size on the board. Do not enlarge the hole size, as this may scratch the resist from the Ground Plane and make an unwanted connection between the lead and 0V. An eye-glass will be found very useful for inspecting connections as they are made, and ensure no unwanted connection is made.

The prototype unit as seen was built using stranded insulated wiring with its ends suitably trimmed and tinned before connection. However, solid core wiring may be a better alternative here.

With all the components in position, carry out the interwiring as shown in Fig. 4. The encircled numerals on some leads

refer to the edge connector fingers they are to connect to. Note that some leads need to be soldered to the board topside.

The socket (slot 3) on board the Apple micro to accommodate the circuitry is shown in Fig. 5.

INSERTING CHIPS

When completely satisfied that the

assembly is complete and correct, insert the i.c.s in their sockets, paying particular attention to their orientation. If these devices are inserted and powered up the "wrong way round", they will undoubtedly be destined for the dustbin.

Initially, set C4 to a mid-way position. It can be set on test if necessary according to results.

COMPONENTS

For Apple II

Resistors

R1	3.9k Ω
R2	3.9k Ω
R3	10k Ω
R4	100 Ω
All	$\frac{1}{4}$ W carbon $\pm 5\%$

See
**Shop
Talk**
page 281

Capacitors

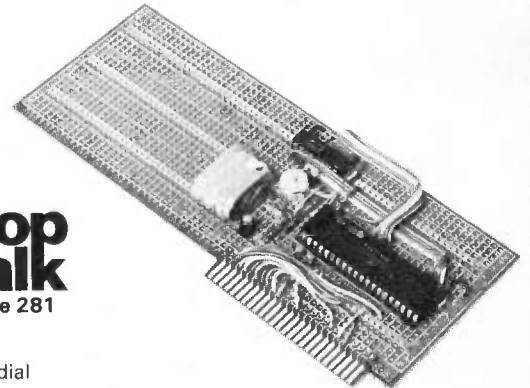
C1	4.7 μ F 6V elect. radial leads
C2, C3	0.1 μ F ceramic or plastic
C4	65pF beehive trimmer capacitor, p.c.b. type

Semiconductors

D1, D2	OA91 or similar germanium diode (2 off)
IC1	MC6821 Peripheral Interface Adaptor (PIA)
IC2	MM58174 micro-processor real-time clock

Miscellaneous

B1	3.6V rechargeable Ni-Cad battery
X1	32.768kHz crystal 40-pin d.i.l. socket; 16-pin d.i.l. socket.



COMPONENTS
approximate
cost **£26**
excluding
software

Apple compatible Microboard, Vero type 200-22266B, see Special Offer Coupon on page 304; p.v.c. insulated connecting wire.

Software

T002 cassette: Real-Time Clock (Apple II).

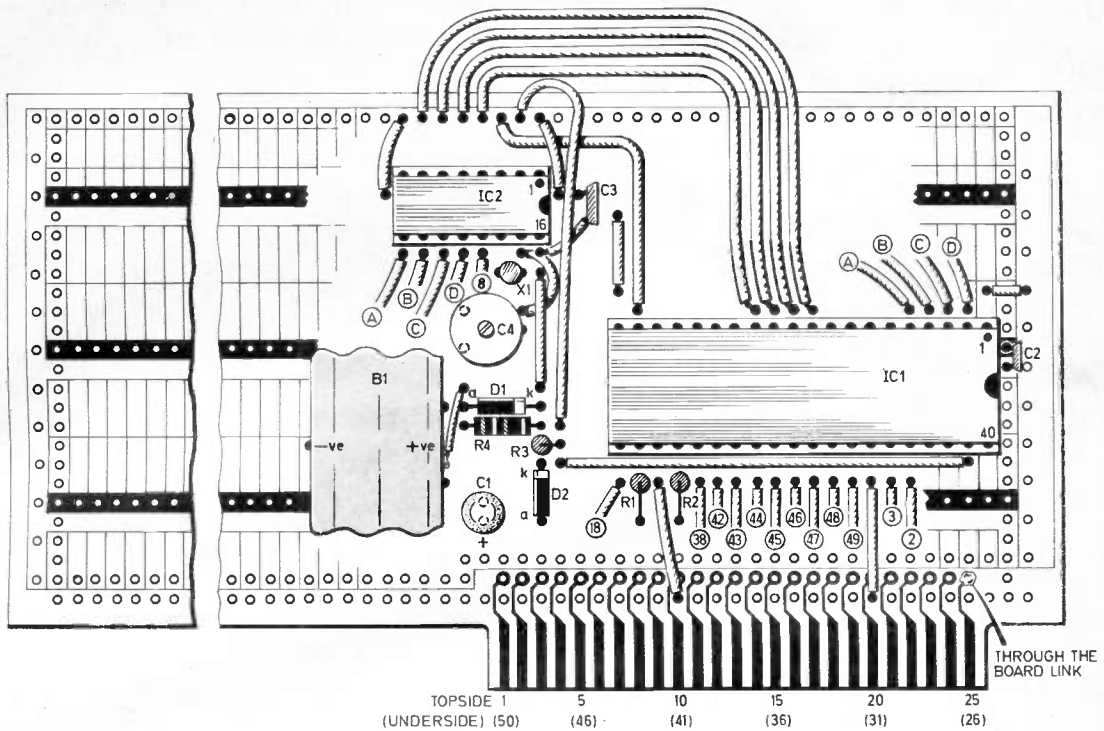


Fig. 4. Topside view of the Apple Compatible Microboard showing component layout and full interwiring. There are no breaks to be made on the underside. Two link wires need to be added on the underside: (1) to connect D1(k) to D2(k) and (2) adjacent to (1) to connect the lead from IC1 pin 39 to the lead from IC2 pin 2. The encircled numbers on each lead refers to the edge connector "finger" they are to connect to.

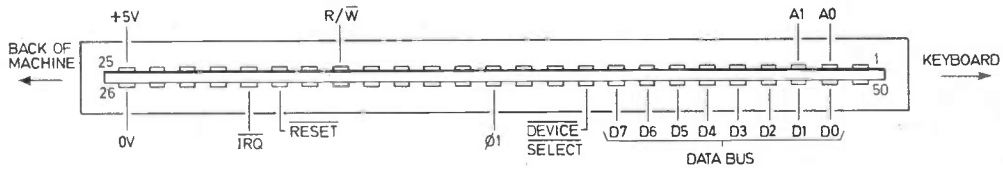
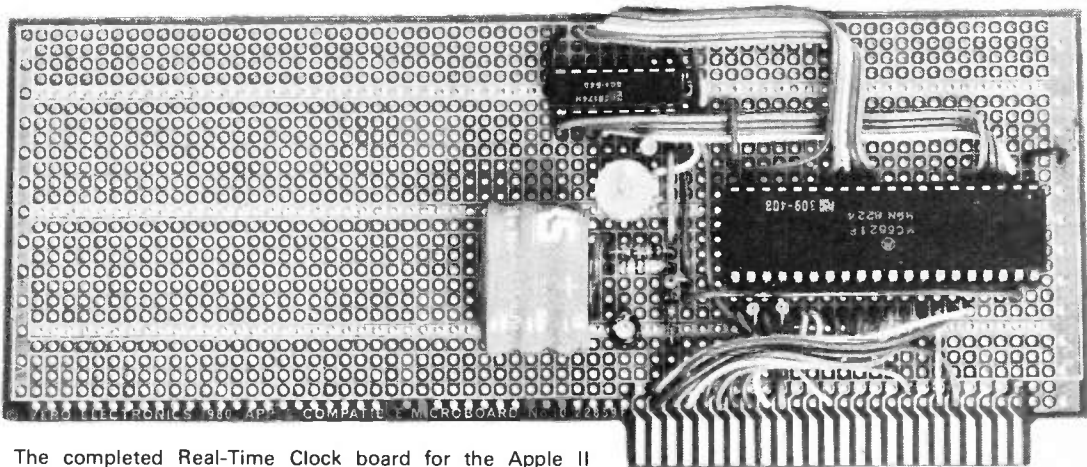


Fig. 5. A "slot" in the Apple II showing the pins required for the Real-Time Clock.



The completed Real-Time Clock board for the Apple II microcomputer.

BBC MICRO REAL-TIME CLOCK CONSTRUCTION



BBC MICRO-BOARD

For use with the BBC Microcomputer, the circuitry is assembled on a single-height Micro-Board. This has a particular arrangement of tracks and supply rails on its underside. This board is to be fitted with a 34-pin plug to allow connection to the BBC 1MHz bus by means of a suitably terminated length of 34-way ribbon cable.

The layout of the components on the topside of the board is shown in Fig. 6. The layout is not critical and may be changed to suit. Sufficient space exists on the board for further user designed circuitry to be added if and when required.

Two breaks are required to be made on the underside of the board, one immediately beneath D1, the other below C1. Make these breaks using a small drill bit and then insert the i.c. sockets. There are numerous link wires required. Some of these were made using tinned copper wire. It was only found necessary to use sleeving in two places, where one wire perpendicularly crosses another, but sleeving could be added at certain places on some wires to prevent the possibility of adjacent wires shorting.

Make all the links and then position and solder the remainder of the components, paying attention to the polarities of B1, D1, D2 and C1. The holes accommodating the leads of B1 and C4 will need to be slightly enlarged to comfortably accept these leads.

It will be found useful to prepare a label to attach to the board immediately behind PL1, containing pin numbering information. This will allow the remaining interwiring to be carried out more easily (and correctly). The wires to reach these locations are marked with an encircled numeral representing their pin destination.

Finally, wire up PL2 and connect its lead to the board. Pinning details for the power outlet into which PL2 is to connect are given in Fig. 7 which also contains the pin numbering for the 1MHz bus.

34-WAY RIBBON CABLE

The geometry of the 34-way ribbon

cable/connectors to comply with the Micro-Board layout in Fig. 6, needs to be as shown in Fig. 8. This configuration must be used. If you already have a cable, but it is wired differently, then to use this cable the wiring to the two rows on the board from PL1 will need to be modified

accordingly.

Set C4 to its mid-way position. After thoroughly inspecting the assembly, and when satisfied that it is correct, the i.c.s may be inserted. Make sure you insert them the right way round, otherwise they will be destroyed when powered up.

COMPONENTS

For BBC Micro

Resistors

R1	3.9k Ω
R2	3.9k Ω
R3	10k Ω
R4	100 Ω
All	$\frac{1}{4}$ W carbon $\pm 5\%$

See
**Shop
Talk**
page 281

Capacitors

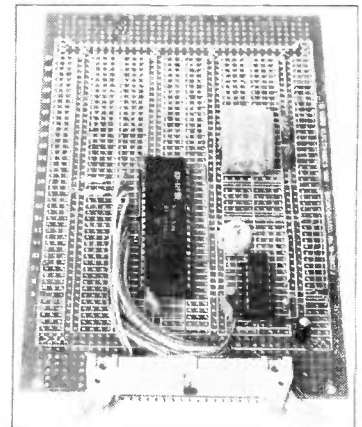
C1	4.7 μ F 6V elect. radial leads
C2, C3	0.1 μ F ceramic or plastic
C4	65pF beehive trimmer capacitor, p.c.b. type

Semiconductors

D1, D2	OA91 or similar germanium diode (2 off)
IC1	MC6821 Peripheral Interface Adaptor (PIA)
IC2	MM58174 micro-processor real-time clock

Miscellaneous

B1	3.6V rechargeable Ni-Cad battery
X1	32.768kHz crystal 40-pin d.i.l. socket; 16-pin d.i.l. socket
PL1	34-way p.c.b. mounting plug with right-angle pins 6-way plug to suit power outlet on BBC Micro
PL2	



COMPONENTS
approximate
cost **£26**
excluding
software

34-way ribbon cable (approx. 50cm) fitted with in-line sockets to mate with PL1 and PL11 on BBC Micro (1MHz bus outlet); tinned copper wire; twin insulated wire for PL2 connection. Microboard, Vero type 200-22271B see Special Offer Coupon, page 304.

Software

T003 cassette: Real-Time Clock (BBC Micro).

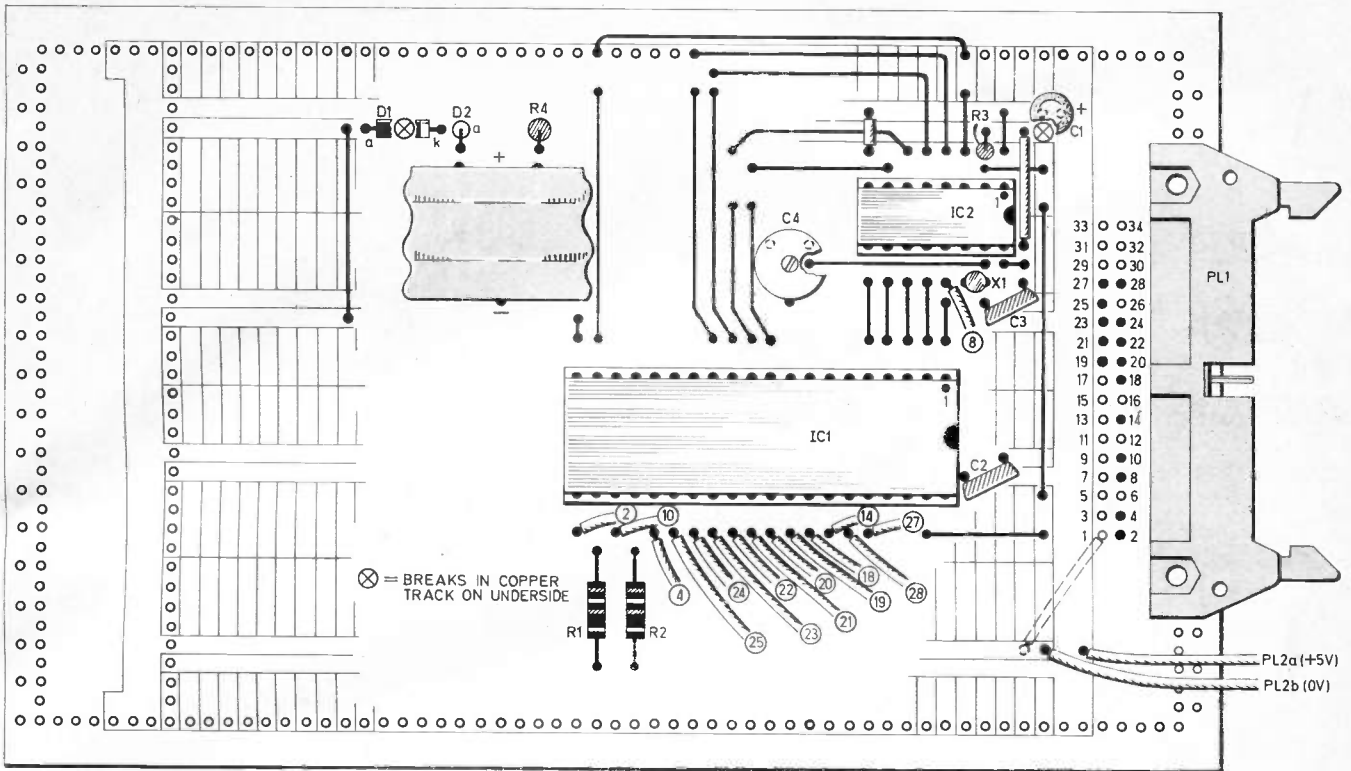


Fig. 6. Topside view of the Micro-Board showing component layout and complete interwiring for the BBC Micro version. The encircled numbers on the leads refer to the 1MHz bus pin no. they are to connect to. The link wire (shown dotted) from PL1 pin 1 to the 0V rail needs to be made if an external power supply (such as the one shown in Figs. 10 and 11), instead of the supply on board the BBC Micro.

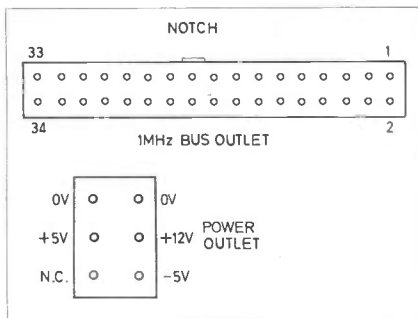


Fig. 7. Socket orientation and pinning information on the BBC Micro as seen when the front of the machine is tipped up and the sockets viewed from the front (keyboard uppermost).

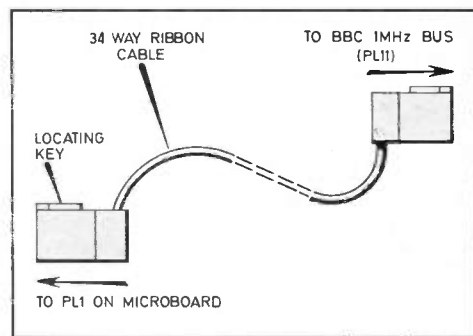
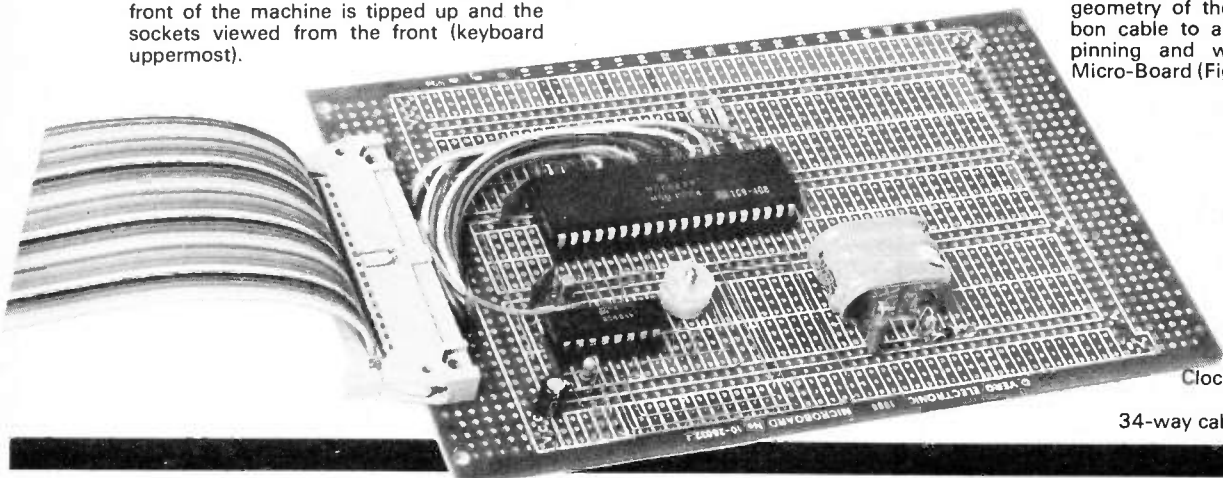
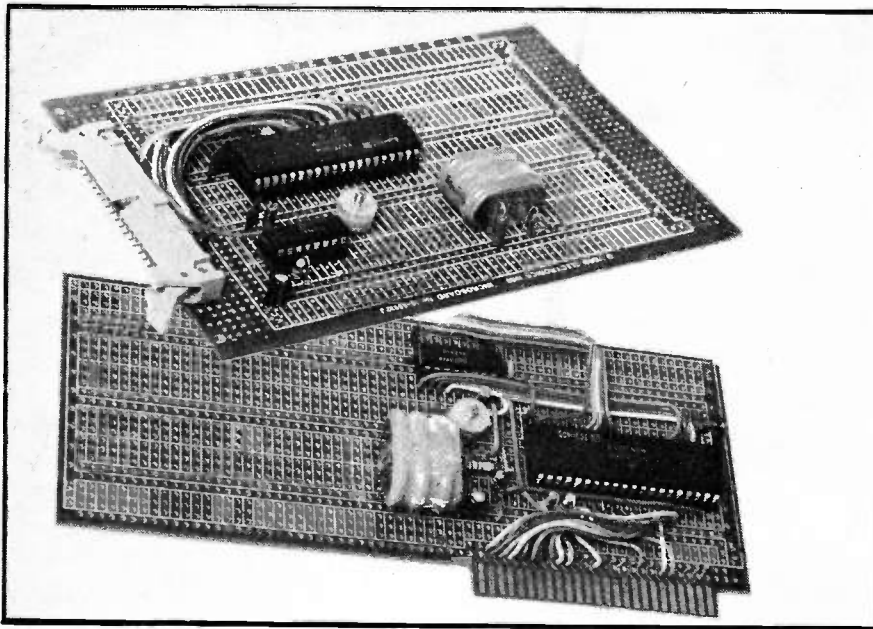


Fig. 8 (above). Required geometry of the 34-way ribbon cable to agree with the pinning and wiring on the Micro-Board (Fig. 6).



The completed Real-Time Clock circuit board Micro with 34-way cable plugged in.



TESTING

The only practicable way of testing it is to plug the circuit board into the computer: into slot 3 on the Apple, or via a 34-way cable/connector into PL11 on the BBC Micro. Switch on the micro, and load in the software.

Since the resetting at power-up causes all registers of the PIA to be set to zero, all lines of the PIA are defined as inputs to begin with. This is a standard safety feature for PIAs, since it prevents the operation of a peripheral such as the clock i.c. from being upset by random output commands at power-up time. In this state, all outputs from Ports A and B and the two Control outputs should be at +5V.

The unused outputs of Port A (pins 6 to 9) remain at +5V throughout, but the unused outputs of Port B (pins 14 to 17) are at 0V since they have no pull-up resistors.

PROGRAMS

The REM statements in the programs indicate what levels to expect on the four used lines of each port and on the control lines.

The INITIALISATION program is used for getting the clock ready for action. It first defines Ports A and B as outputs, then sets register 1 (test) and register 14 (start/stop) to zero. The clock is now ready to be set to the correct time and date.

The TESTING program simply allows you to examine the contents of any clock register, to see if it is behaving as expected. It begins by starting the clock, converts Port B to input function, and then asks you which register you want to examine. You will note that, before it prints the data it has read, it checks to see if it is

"15". A reading of "15" is obviously nonsense as far as times and dates are concerned, and its appearance is a warning from the clock that its registers have been updated since the last reading.

For example, suppose the time is read just before 4 o'clock. The micro might read 59.9 seconds and 59 minutes, but before it can read the hours the clock changes to 4 hours 0 mins 00 seconds. The micro will read the hours as 4, and print out the time as 4 hours 59 minutes 59.9 seconds, an error of almost 1 hour! Updating of any register causes it to output "15", which means that you should begin the process of reading the registers again. There is plenty of time to read them all in a tenth of a second.

If you run the program and answer <3> in response to the query, you should see a column of single digits scrolling up the screen, incrementing by 1 every second. If you select register <4> instead, they should increment every ten seconds.

The SETTING program allows you to set the time while the clock is stopped. Set it for a few minutes ahead of the actual time. Wait when the message "Ready?" is

printed. As soon as the time is exactly that which you have set, press <Y>. This starts the clock at precisely the right instant.

When you have tried the TESTING program and examined the way it works, you will be ready to write your own program for reading the time.

READING shows you how to read hours and minutes and the day of the week. You can extend the program to take in seconds, and the date.

Having read the data, there are all kinds of ways of displaying it, from a straightforward text display to a simulation of an analogue clock dial, with moving seconds hand. There is quite a lot of programming fun to be had from the real-time clock.

INTERRUPTING

The "Interrupt" facility needs a machine code program. If you do not wish to become involved in machine code programming you may leave out the signal line connecting the interrupt output of the clock (pin 13) to the \overline{IRQ} line of the computer.

Normally the \overline{IRQ} line is held at high level, but the clock can bring it low to cause an Interrupt ReQuest. When this happens, and provided that the microprocessor has been enabled to respond to such a request, it completes the operation it is engaged in, then stores the contents of its registers in a safe place, the "Stack".

It then goes to addresses 1022 and 1023 in RAM, where it expects to find the starting address of its Interrupt Service Routine. The Interrupt Service Routine (ISR) tells the MPU what to do next.

If, for example, the clock is being used in conjunction with a games program to limit the length of time each player is allowed to take for each move, the ISR might cause the message "Time up! Next player's move" to be displayed. When the ISR is completed, the microprocessor jumps back to the point it had reached in the main program, retrieves the original contents of its registers from the Stack and continues with the main program.

An interrupt does not necessarily have any apparent effect on the program which is being run. An egg-timer program could

Table 4. Interrupts: data to be placed in Register 15

Function		Bit 3210	Decimal equivalent
No interrupt		0000	0
Single interrupt	after 60s	0100	4
	after 5s	0010	2
	after 0.5s	0001	1
Repeated interrupts	every 60s	1100	12
	every 5s	1010	10
	every 0.5s	1001	9

be run at the same time as any other program. If the clock is set to interrupt at 1-minute intervals, the microprocessor is interrupted, jumps to a routine to count the number of times it has been interrupted and returns to the main program. It happens so quickly that you would not notice the interruption.

On, say, the fifth interruption, when the egg is ready for eating, a message "Egg ready now" is flashed on the screen, before the main program is resumed.

INTERRUPT FLAG

\overline{IRQ} is enabled if the Interrupt Flag in the Status Register of the microprocessor is "0", and disabled if it is "1". When the computer is first switched on, this Flag is set to "0". Unless you deliberately set the flag to "1" by using SEI in machine code, it remains at "0" so that there is no need to take special action to enable it.

The only time it goes to "1" of its own accord is when the microprocessor is servicing an interrupt and does not want to be interrupted while doing so. When it has finished it makes the flag "0" again.

To tell the microprocessor where to find the start of the ISR, you place the starting address in locations 1022 and

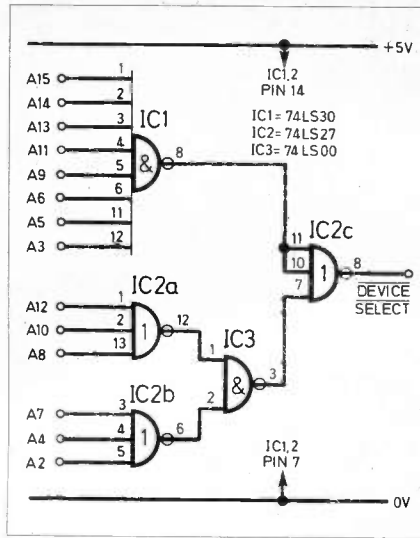


Fig. 9. Address decoder for the addresses 60008 to 60011. Lines A0 and A1 go to RSO and RS1 of IC1 in Fig. 2.

1023. The low byte of the address goes in 1022, the high byte in 1023.

The address can be changed from time to time during the course of a program so

that the interrupt sends the microprocessor to different ISRs which cause different things to happen on at different stages of the program. This can be done simply from a basic program by POKE-ing the new addresses to 1022 and 1023.

The interrupt output of the clock is controlled by Register 15. The settings are shown in Table 4. The procedure for setting the register is exactly the same as that used in the SETTING program for the other registers. Having set Register 15 to the required interval, read the Register three times in succession. This starts the interrupt timer.

OTHER MICROS

Unless your micro has "slots" or a built-in I/O port, you will need to provide an address decoder for the clock. You can choose any address you like, but take care that it is not one which is already allocated for use by the micro itself.

For example, it must not be in the range used by RAM, ROM, the Keyboard or the Video monitor. Consult your manual for details. Having found a vacant address, an address decoder can be built up on the model of that shown in Fig. 9.

5V POWER SUPPLY

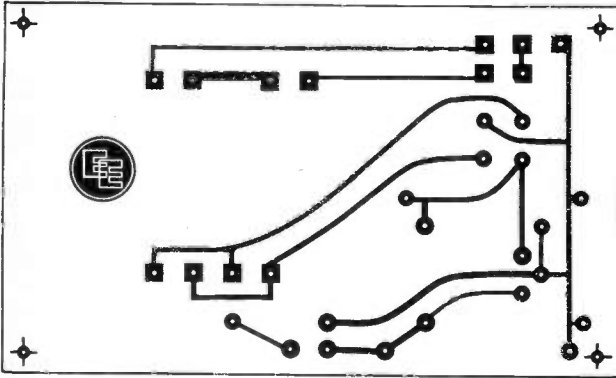


Fig. 11. The 5V Regulated Power Supply is assembled on one p.c.b., the full size artwork of which is shown here. It must be housed in a suitable case and the output taken to a pair of sockets mounted alongside the switch and i.e.d.

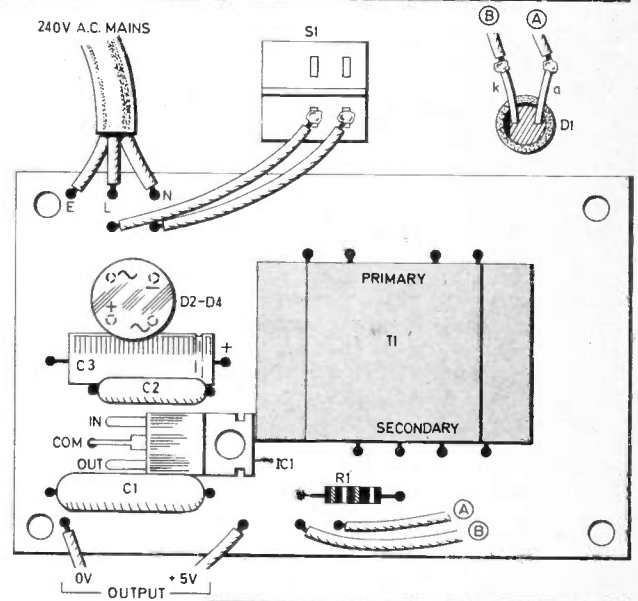
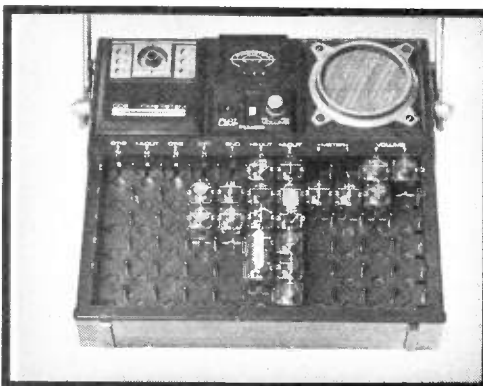
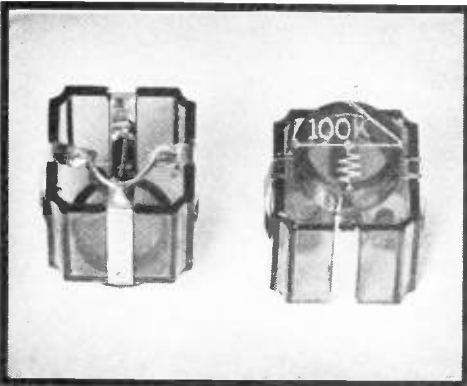


Fig. 10. Circuit diagram of the 5V Regulated Power Supply.

COMPONENTS

R1	470Ω ¼W carbon ±5%
C1	470nF polyester C280
C2	220nF polyester C280
C3	220µF 10V elect.
D1	TIL220 0.2in red i.e.d.
D2-5	W005 50V, 1A bridge
IC1	7805 5V, 1A regulator
S1	s.p.s.t. mains switch
SK1,2	4mm banana socket (2 off)
T1	mains transformer with 0-6V, 0-6V 250mA secondaries, p.c.b. mounting (RS 207-829)

Single-sided p.c.b. 80 x 50mm; case to suit; wire; mains lead.



SPECIAL REPORT

FX — COME

FIRST impressions of the FX Computer are that it is a very deceptive piece of equipment. It may look like a cheap toy but neither is it cheap nor is it a toy.

Built by Gakken Corporation Limited, this educational kit retails at £69.95 plus £3.00 for p & p, available from Electroni-Kit Limited, Dept. EE, 388 St. John Street, London, EC1V 4NN.

The case is made from robust plastic and forms the experimenters "test-bed". The top section of the case near the carrying handle contains the supply on/off switch, the amplifier module, volume control and light sensitive cell (CDS). The CDS module may be interchanged for an i.e.d. clock and timer.

TEST-BED

The test-bed is located in the lower section of the case and is designed around a rectangular well which accepts the various component cubes. The base of the well is marked up into a matrix of squares. The component cubes are located in these squares, held in position by small plastic locating pillars at the corners of the squares.

The sides of the well connect to the component cubes via sprung metal connecting lugs. The contacts on the sides of the well are wired to various modules mounted in the top section of the case. These well contacts are clearly marked, so no mistakes should be made when constructing the electronic circuits.

COMPONENT CUBES

The kit provides 29 component cubes made from preformed plastic. The various components are mounted in the centres of the cubes which are clearly marked with the component symbol and value. The cubes also have sprung connecting lugs on their sides which in turn are wired to the components. Cubes without components are used as interwiring links.

Far left shows the component cubes with symbol and component mounting. (Centre) The test-bed with an experimental Gun Sound circuit. (Left) Test-bed showing the computer keyboard ready for use.

KEYBOARD

The computing circuitry of the kit is contained in the keyboard, which in turn slots into the test-bed. The keyboard is then able to draw power via connecting lugs. The keyboard uses a custom designed chip to provide the program memory, data memory and registers. The keyboard is a standard hexadecimal key located board and is neat and attractive. Both binary and hexadecimal displays are located on the board, and these provide clear viewing when programming.

Certain keys are designated a musical note and a full range musical scale is available. The unit provides reasonable sound and the notes are of good clarity.

Various games are already stored permanently in the memory and are used in conjunction with the binary display.

INSTRUCTION MANUALS

Provided with the kit are two instruction manuals. One manual deals with the computing side of the kit and lists 100 example programs. The second manual handles the electronic projects and provides 65 example circuits.

The electronic projects manual is excellently illustrated and each project has its own circuit diagram and easy to follow "building block" component layout. A brief introduction to each circuit is also provided.

The computer manual is not as clear as might be expected, especially when dealing with computing in Machine Code. The manual is also confusingly ordered and it lacks any real continuity.

CIRCUIT AND PROGRAM TEST

All 165 computer and electronic projects were carried out.

We found the electronic circuits easy to set up and nearly all the circuits were successful. Those that failed were due to an error on our part. The Steam Boat Sound circuit provided the first failure but after re-checking the circuit, it was found to be due to a component cube making an unnecessary connection with the wall of the well.

We found that it was possible to make

KIT CONTENTS

1	Test Bed
1	Computer Console
1	Light Sensitive Cell
1	Speaker
29	Component Cubes
2	Test Probes
1	Manual containing 65 electronic projects
1	Manual containing 100 computer projects

a bad connection when using a small amount of component cubes. This was because the cubes lacked any pressure to push them together. It may be a good idea to push each of the cubes firmly down once the circuit has been built.

The computer programs of which there are 100 examples proved to be just as successful. Few mistakes were made when operating the keyboard, which proved to be very fast and reliable.

Any mistakes that were made when programming were easily edited, although at times it was easier to "reset" and start again.

CONCLUSION

The kit is very well made and the layout of the unit is excellent, the idea of a test-bed may not be new but it provides an easy circuit set-up.

Electronic projects are easily set up in the test-bed and if successful the circuit may then be drawn from the symbols printed on the cubes. This application is extremely useful, even to an experienced electronics engineer.

The kit has good educational value and the electronic project manual provides a first class introduction for the beginner. The computer manual provides a good introduction to computing, but then tends to lose its way and is difficult to follow.

The components in the cubes are easily interchangeable, so that the kit may be easily expanded, to a limitless number of projects.

Overall a good educational kit, although the price is a bit questionable. □

COMPUTER KIT

BY R. HOOPER

previously passed through a longer piece of plastic tube. The joints were sleeved. The two parts of plastic tubing clipped together.

The LM335 can be mounted in a thin metal tube to provide protection for it. The metal cases often used for clinical thermometers are ideal, although some heatsink compound should be used to improve thermal contact between sensor and casing avoiding electrical contact between the two.

Connect the battery, insert in the case and fit the case lid to complete the unit.

A digital voltmeter is required for accurate calibration. With the LM335 at 0 degrees Celsius, connect a voltmeter across the output (PL1) and adjust VR1 for ZERO output. The unit is then ready for use with the A-D Converter.

SOFTWARE

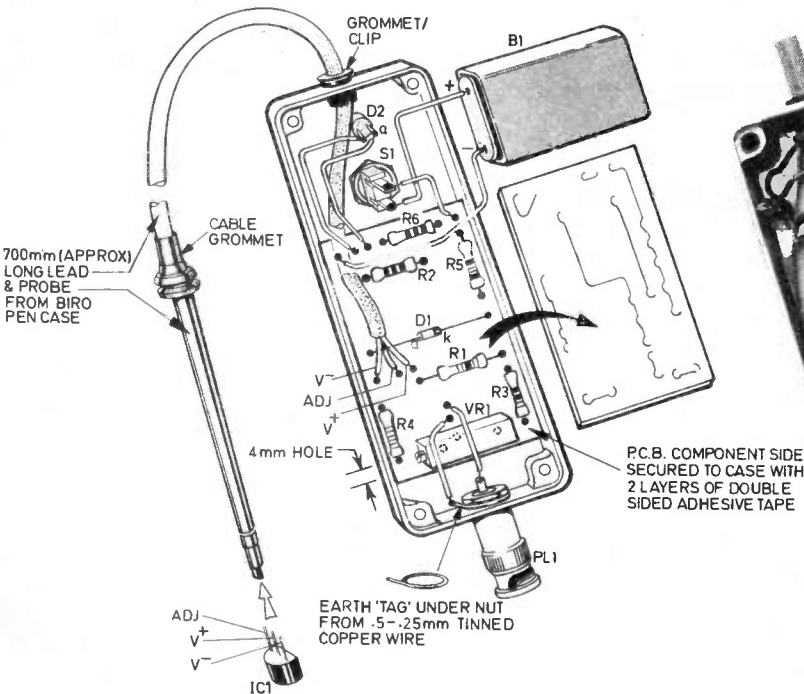
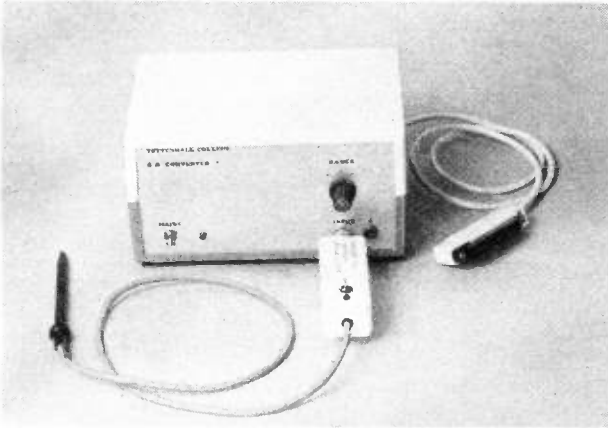
High level languages are not generally well suited to input/output (I/O) control applications: writing a peripheral interfacing program in Basic can be likened to attempting to eat an apple through the strings of a tennis racket.

It is possible, however, to marry the convenience of Basic with the efficiency of machine code by patching a machine code program into the PET Basic character fetch routine (CHRGET) in page zero. This enables the programmer to utilise the inherent benefits of a machine code subroutine to control the operation of the A-D Converter without sacrificing the advantages of Basic.

The fact that the entire operation of such a converter can be controlled by a

The completed prototype sensor plugged into the A-D Converter. The latter is suitable for use with PET and Vic 20.

A view inside the completed prototype sensor. The board is held securely by double-sided foam on its underside (see below).



700mm (APPROX)
LONG LEAD
& PROBE
FROM BIRO
PEN CASE

EARTH TAG UNDER NUT
FROM .25mm TINNED
COPPER WIRE

P.C.B. COMPONENT SIDE
SECURED TO CASE WITH
2 LAYERS OF DOUBLE
SIDED ADHESIVE TAPE

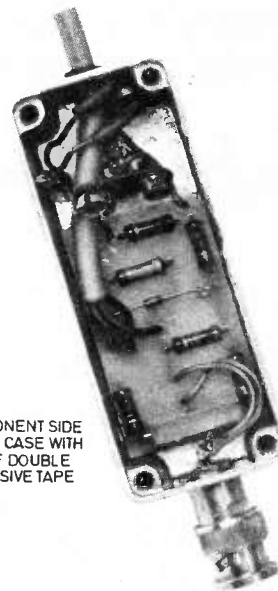


Fig. 3. Layout of the components on the p.c.b. topside with full assembly and interwiring details.

COMPONENTS

Resistors

R1	10kΩ
R2	4.7kΩ
R3	3.9kΩ
R4	2.2kΩ
R5	2.2kΩ
R6	270Ω
All ¼W carbon ±5%	

See
**Shop
Talk**
page 281

Semiconductors

IC1	LM335Z precision temperature sensor
D1	BZY88C8V2 8.2V 400mW Zener diode
D2	TIL220 0.2 inch red l.e.d.

Miscellaneous

S1	miniature on-off toggle
PL1	b.n.c. plug
B1	PP3 9 volt
VR1	1kΩ ¼ inch multiturn cermet preset

Printed circuit board, single-sided size 51 x 30mm; 3-core cable; diecast aluminium box, size 90 x 35 x 30mm; PP3 battery clip; sleeving; rubber grommet.

Approx. cost
Guidance only **£6**

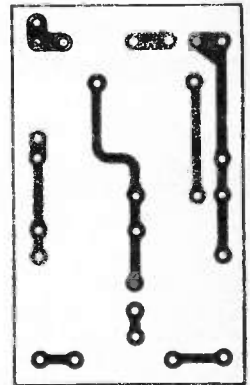


Fig. 2. Full-size master p.c.b. pattern. The black areas represent the copper to remain after etching.

single command in a Basic program means that experimental results are easily obtainable for use within that program.

APPLICATIONS

Some applications in the chemistry laboratory are outlined below. In each case a temperature probe is attached to the converter.

1—DIRECT MEASUREMENT

The FETCH command, !T, instructs the converter to change the analogue input into a digital value which is passed to the Basic program as variable T.

```
10 !T
20 PRINT "PROBE TEMPERATURE IS"; T+273; "KELVIN"
```

2—DATA LOGGING

The FETCH command, !T, can be placed in a timed program loop. Temperature readings can thus be stored at regular intervals in a dimensioned array. One such sampling program is listed below.

```
10 DIM T(100)
20 DELAY = 600: REM 10 SECONDS
30 FOR N = 1 TO 100
40 S = TI
50 IF TI < S + DELAY THEN 50
60 !T
70 T(N) = T
80 NEXT N
```

3—DIGITAL DISPLAY

The temperature obtained using the FETCH command, !T, can be passed to

a number plotting sub-routines: the large digits so displayed can be seen by all pupils in the laboratory.

```
10 !T
20 GOSUB 10000: REM NUMBER PLOTTING ROUTINE
```

4—GRAPH PLOTTING

A number of different graphical formats are possible. If quarter character graphics are used in place of the normal full character graphics a reasonable 80x50 screen resolution can be obtained. A direct graphical representation of a temperature/time curve during the cooling of naphthalene is one such application of this technique.

5—DECISION MAKING

The use of FETCH commands and IF... THEN statements in a program allow decisions to be made. One obvious application is in boiling point determination—is the temperature increasing or has it reached a constant value?

6—COMPUTER SIMULATION

A stainless steel probe containing a platinum resistance thermometer and hydrogen gas can be interfaced via the converter to the microcomputer. The temperature of the gas is monitored and a bar chart of the Maxwellian distribution of molecular velocities at different temperatures is plotted during immersion in various liquids.

PROGRAM A FOR PET COMPUTER

```
5 REM TEMPERATURE/PET/ZN427E
9 REM INITIALISATION ROUTINE
10 POKE 59468,PEEK(59468)OR192:REM CB2 ENABLE
20 POKE 59468,PEEK(59468)AND223:REM TRI-STATE DISABLE
30 POKE 59456,PEEK(59456)OR8:REM START CONVERSION HIGH
40 POKE 59459,0:REM PORT A INPUT
50 GOSUB 2000
60 PRINT "TEMPERATURE";TCX
70 GOTO 50
1999 REM CONVERSION ROUTINE
2000 POKE 59456,PEEK(59456)AND247:REM START CONVERSION LOW
2010 POKE 59456,PEEK(59456)OR8:REM START CONVERSION HIGH
2020 POKE 59468,PEEK(59468)OR32:REM TRI-STATE ENABLE
2030 TCX=INT(PEEK(59471)*.4+.5):REM FETCH TEMPERATURE
2040 POKE 59468,PEEK(59468)AND223:REM TRI-STATE DISABLE
2050 RETURN
```

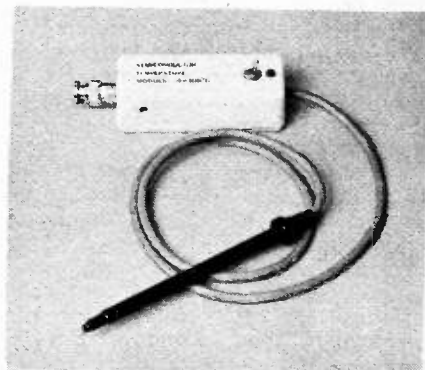
PROGRAM B FOR VIC 20 COMPUTER

```
5 REM TEMPERATURE/VIC-20/ZN427E
9 REM INITIALISATION ROUTINE
10 POKE 37147,PEEK(37147)AND227:REM DISABLE SHIFT REGISTER
20 POKE 37148,PEEK(37148)OR192:REM CB2 MANUAL OUTPUT
30 POKE 37148,PEEK(37148)AND223:REM TRI-STATE DISABLE
40 POKE 37139,PEEK(37139)OR32:REM DORA BIT 5 OUTPUT
50 POKE 37151,PEEK(37151)OR32:REM START CONVERSION HIGH
60 POKE 37138,0:REM DORB INPUT
70 GOSUB 2000
80 PRINT "TEMPERATURE";TCX
90 GOTO 70
1999 REM CONVERSION ROUTINE
2000 POKE 37151,PEEK(37151)AND223:REM START CONVERSION LOW
2010 POKE 37151,PEEK(37151)OR32:REM START CONVERSION HIGH
2020 POKE 37148,PEEK(37148)OR32:REM TRI-STATE ENABLE
2030 TCX=INT(PEEK(37136)*.4+.5):REM FETCH TEMPERATURE
2040 POKE 37148,PEEK(37148)AND223:REM TRI-STATE DISABLE
2050 RETURN
```

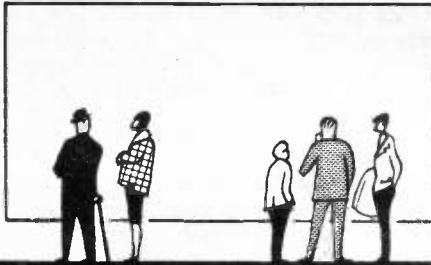
TRANSDUCERS

As previously mentioned, existing laboratory measuring instruments which have an "external meter" connection can be interfaced with the microcomputer using the Converter. The 100mV range is often useful here although the instrument should be provided with a "dummy" load resistor of value equal to the resistance of the meter with which it is normally intended to be used. (Many such meters are 100Ω, 1mA f.s.d. types and a resistor of this value is thus suitable.)

Tailor made transducers are under development by the designers and include the following: temperature (semiconductor), temperature (platinum resistance), pH value, luminous intensity, sound level and these designs will be published in EE in due course. □



SHOP TALK



BY DAVE BARRINGTON

Catalogues

This month we have received two new mail order catalogues from firms which, until now, are new to us.

The 132-page Electronic Components Catalogue from **MS Components** is lavishly illustrated and every item is priced under each entry. Over 1800 new items have been added to their already vast stock of components.

Components held range from resistors and capacitors to computer keyboard and printer mechanisms. A new 33-page section features a broad range of discrete components, logic, linear, hybrid, microprocessor, interface and memory integrated circuits.

Although supplying mainly to the trade, MS Components do *not* levy a minimum order charge and prices include postal charges. Also, it states under their conditions of sale that they "will despatch to any type of customer or private individual provided that the terms of our trading conditions are compiled with".

The MS Components catalogue is well worth adding to readers' libraries and further details and copies may be obtained from **MS Components Ltd., Dept EE, Zephyr House, Waring Street, West Norwood, London, SE27 9LH.**

Two items that caught the eye whilst browsing through the new 176-page **JEE Distribution Catalogue** were a range of board and cable fasteners and a range of pinout labels for the popular 7400 and 4000 series of integrated circuits. The catalogue is issued with a separate price list which is updated throughout the year.

The full-size self-adhesive labels have the relevant pinning details, including type number, printed on them and are stuck on top of the i.c. to give at-a-glance information of the internal make-up of the device.

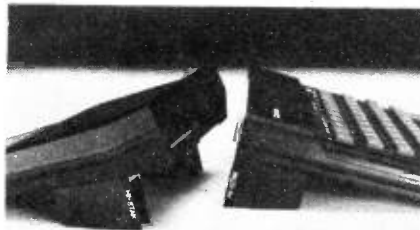
Coinciding with the release of the JEE catalogue is the announcement of the opening of their new shop premises in Southfields, London.

Copies of the JEE Distribution Catalogue can be obtained from **JEE Distribution Ltd., Dept EE, 43 Strathville Road, London, SW18 4QX.**

Home Computer Feet

Claiming to make keyboards more comfortable to use and their printed surfaces easier to read, **Warp Factor Eight** have introduced plastics "ramps" for raising home computers at a more inclined angle.

Known as Hi-Stak they are moulded plastics ramps with built-in rubber feet and self-adhesive tops which are stuck to the rear of machines to raise the rear-end from the resting surface. The additional "feet" are suitable for Sinclair, VIC20, Atom, Jupiter Ace, Lynx and Oric 1.



Warp Factor Eight Hi-Stak home computer feet.

The Hi-Stak is available, by mail order, from **Warp Factor Eight, 6 Pelham Road, Braughing, Ware, Herts.** Price £3.95 including VAT and postage.

Spectrum Sounds

Three months ago **Bi-Pak Semiconductors** marketed a sound generator for use with the Sinclair ZX81 (See *Shoptalk*, February issue).

EVERYDAY ELECTRONICS SOFTWARE SERVICE

The EE Software Service provides an easy and reliable means of program entry for our computer-based projects. All programs have been tested by us and consist of two good quality copies of the working program on cassette tape.

So successful has this been that they have now introduced a new modified version for all the Sinclair range of home computers.

Designated ZON X, the unit is self-contained in a black plastics case with a loudspeaker and volume control. For use with the Sinclair Spectrum, there is a further plug-in adaptor which houses a crystal and other components needed to give unlimited sound facilities.

The ZON X for use with ZX81 and Timex 1000 cost £25.95. The ZON X plus special adaptor for the Spectrum cost £32.75. All prices include VAT and postage.

Further details may be obtained from: **Bi-Pak Semiconductors, Dept EE, The Maltings, 63a High Street, Ware, Herts, SG12 9AG.**

CONSTRUCTIONAL PROJECTS

Headphone Amplifier

When ordering the input jack socket SK 1 for the *Headphone Amplifier*, be sure to specify a d.p.d.t. switched type. These are stocked by most of our advertisers.

MW Radio/Tuner

The ferrite aerial for the *MW Radio/Tuner* is available direct from Denco (Clacton) Ltd., Dept EE, 355 Old Road, Clacton-on-Sea, Essex. Other suitable aerials are stocked by Ambit and Maplin.

Model Train Controller

The case used in the prototype model of the *Model Train Controller* was obtained from Maplin and can be ordered as: LH67X (ABS Console M6007). Most of our advertisers stock sloping front cases and any of these may be used, provided they are of similar dimensions.

Real-time Clock

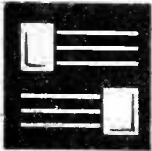
Quite a large saving on the circuit boards for the Real-time Clock projects can be obtained by using the special coupon on page 304. The coupon should be presented to your local BICC-Vero stockist.

The only stockist we have been able to locate for the 6-way plug to suit power outlet on the BBC Micro (PL) is Watford Electronics, Dept EE, 34 Cardiff Road, Watford, Herts, WD1 8ED. We understand that they are able to supply all components for this project.

The semiconductors, crystal, d.i.l. sockets and connectors (except the 6-way plug) are available from Cricklewood Electronics.

PROJECT TITLE	CODE	COST
ZX81 SPEED COMPUTING SYSTEM (Feb 83)	T001	£2.95
REAL-TIME CLOCK (Apple) (May 83)	T002	£2.95
REAL-TIME CLOCK (BBC) (May 83)	T003	£2.95

GUITAR HEADPHONE AMPLIFIER



A SIMPLE PROJECT FOR YOUR FREE TRANSISTORS

BCY65EP—2N4123

It is possible to practice playing the electric guitar without causing any disturbance or annoyance to others by simply relying on the acoustic output of the instrument, but this has the drawback of very low volume and it is obviously impossible to practice using any guitar effect units.

A much better alternative is to use a practice amplifier having a headphone output, or a headphone amplifier such as the simple and inexpensive design featured here.

The circuit is primarily intended for use with high or medium impedance headphones, but it also seems to give quite good results when used with low cost 8 ohm impedance headphones. The sensitivity of the amplifier is high enough for use with low output guitar pick-ups, but a volume control at the input of the circuit enables the sensitivity to be reduced so that overloading can be avoided if a high output pick-up is used. The unit also has proper bass and treble tone controls.

CIRCUIT DESCRIPTION

The circuit has just two stages; a preamplifier built around TR1 and an active tone control stage based on TR2. The full circuit diagram of the unit is shown in Fig. 1.

TR1 is used as a common emitter amplifier with R2 as the collector load resistor and R1 providing base biasing. R3 is used to provide negative feedback which boosts the input impedance of the circuit slightly, as well as giving lower noise and distortion levels. C3 reduces the response of the preamplifier at frequencies above the audio range, and this helps to avoid instability and breakthrough of radio frequency interference. The input signal is coupled to TR1 via volume control VR1 and d.c. blocking capacitor C2.

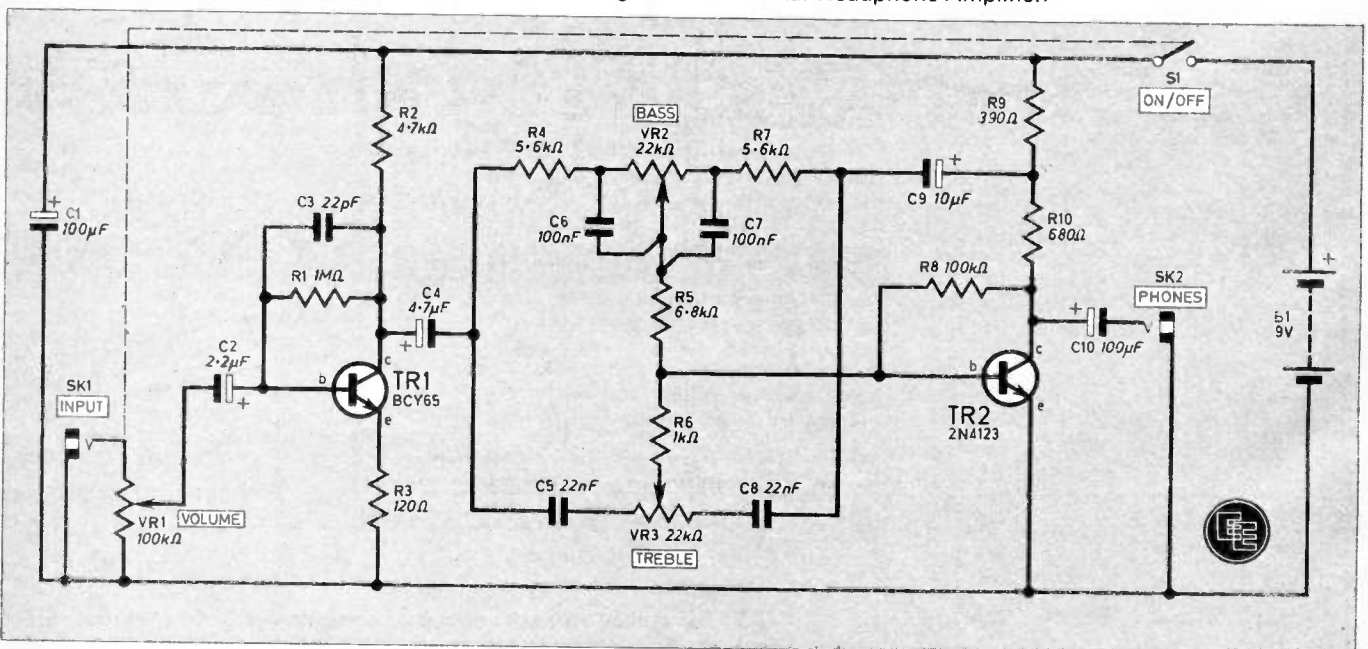
The tone control circuit uses TR2 as a common emitter amplifier with frequency selective negative feedback provided via the tone control networks. The circuit operates in a similar manner to an inverting mode operational amplifier circuit, and the voltage gain of TR2 is roughly equal to the impedance from C9 to the base of TR2 divided by the impedance from C4 to the base of TR2.

TREBLE CONTROL

If we consider the treble control circuit first, the relevant feedback components are C5, C8 and VR3. At low frequencies C5 and C8 will have impedances that are high in comparison to the track impedance of VR3, and while adjustment of VR3 will obviously have some effect on the gain of the circuit, the degree of control available is insignificant.

The situation is very different at high frequencies where C5 and C8 have a very low impedance. With the wiper of VR3 towards the C5 end of its track there is a low impedance from C4 to the base of

Fig. 1. Complete circuit diagram of the Guitar Headphone Amplifier.



TR2, and a much higher impedance from C9 to the base of TR2.

The circuit consequently provides a reasonable voltage gain at these frequencies. With the slider of VR3 towards the C8 end of its track there is a low impedance from C9 to TR2's base, and a high impedance from C4 to TR2's base. The circuit consequently provides less than unity voltage gain.

Thus VR3 provides the required effect by providing boost and cut at treble frequencies, but having no significant effect at bass and middle frequencies.

BASS CONTROL

The bass control (VR2) operates in a similar fashion. At high frequencies C6 and C7 place what is effectively a short circuit across VR2 so that it can have no significant influence on the gain of the circuit. At low frequencies the impedance of C6 and C7 is much higher and VR2 is then able to exercise considerable control over the gain of the circuit. R4 and R7 are included to prevent excessive bass boost and cut being provided by the circuit. R5 and R6 help to minimise interaction between the bass and treble tone control networks.

In this explanation it has been assumed that the feedback is taken direct from the output of TR2 (collector terminal), but it is in fact taken from a tapping on the collector load resistance. This does not fundamentally affect the operation of the tone controls, and merely results in a small voltage gain through TR2, rather than a nominal unity voltage gain.

This small additional voltage gain is needed to ensure that the amplifier has adequate gain for use with low output guitar pick-ups which provide an output signal of only about 20 or 30mV r.m.s. in amplitude.

S1 is the on/off switch and is a set of make contacts on the input socket, SK1. The unit is therefore automatically switched on when the guitar is connected to SK1, and switched off again when it is disconnected. A separate on/off switch can obviously be used if preferred. The current consumption of the circuit is about 5mA, and power is provided by a 9-volt battery.



CASE

An aluminium box measuring about 133 x 102 x 38mm provides a housing for the unit which is reasonably compact but does not complicate construction. The three controls and the input socket are mounted on the front panel, and wiring the unit will be more straightforward if the panel layout used on the prototype is copied. In other words these components should be mounted in a single row

COMPONENTS

Resistors

R1	1M Ω	R5	6.8k Ω
R2	4.7k Ω	R6	1k Ω
R3	120 Ω	R8	100k Ω
R4,7	5.6k Ω (2 off)	R9	390 Ω
		R10	680 Ω

All $\frac{1}{4}$ W carbon $\pm 5\%$

Capacitors

C1	100 μ F 10V axial elect.
C2	2.2 μ F 63V radial elect.
C3	22pF ceramic
C4	4.7 μ F 63V axial elect.
C5,8	22nF polyester (2 off)
C6,7	100nF mylar (2 off)
C9	10 μ F 25V axial elect.
C10	100 μ F 10V radial elect.

Semiconductors

TR1	BCY65EP <i>n</i> pn silicon
TR2	2N4123 <i>n</i> pn silicon

Miscellaneous

VR1	100k Ω log carbon potentiometer
VR2,3	22k Ω lin carbon potentiometer (2 off)
S1	Part of SK1
SK1	Standard jack with d.p.d.t. contacts
SK2	Standard stereo jack
B1	9 volt (PP3 or PP6 size) 0.1in matrix stripboard, 20 strips by 24 holes; aluminium case, 133 x 102 x 38mm (type AB10); three control knobs; battery connector; wire; 6BA fixings.

Approx. cost
Guidance only **£5.00**

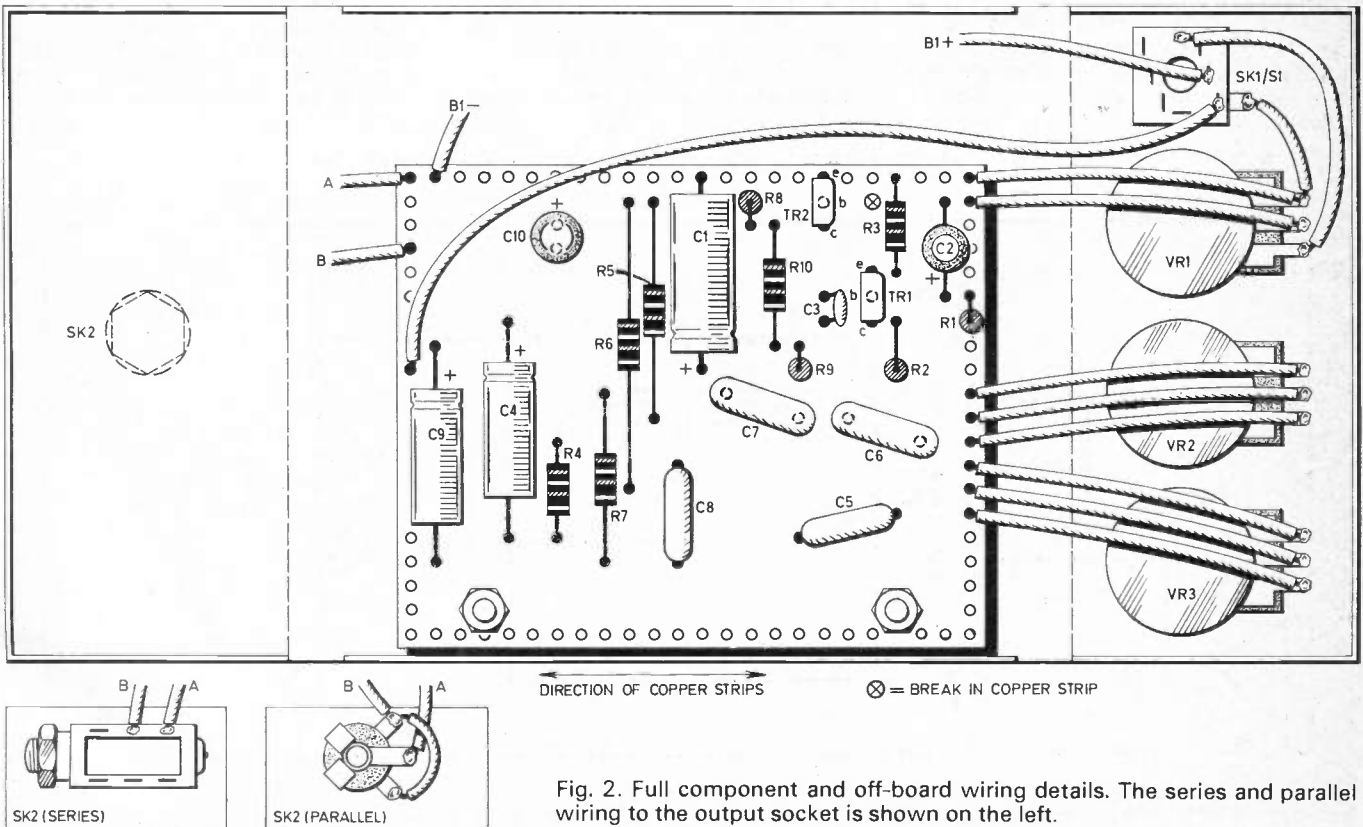
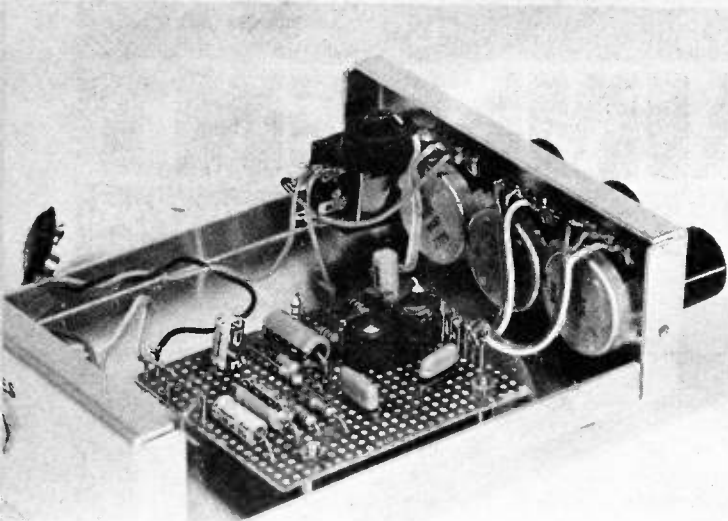
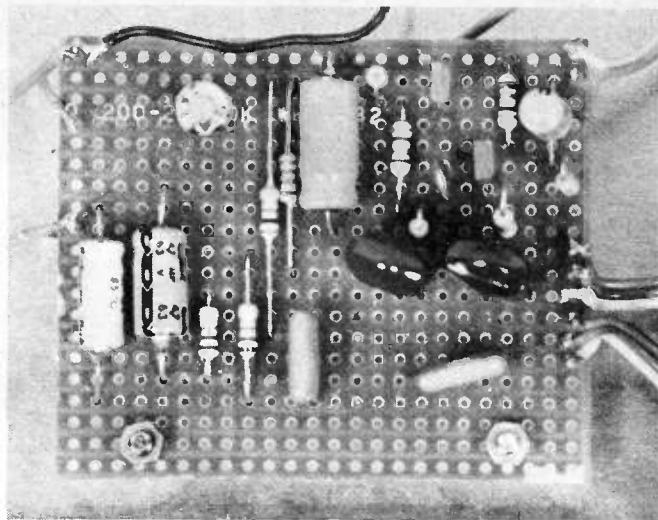


Fig. 2. Full component and off-board wiring details. The series and parallel wiring to the output socket is shown on the left.



Picture showing layout of components inside the case.



Component board layout.

with VR3 on the left, then VR2, next VR1 and SK1 on the right. SK2 is fitted on the rear panel of the amplifier.

COMPONENT BOARD

The component panel is a 0.1 inch matrix stripboard panel which measures 20 copper strips by 24 holes. Details of the component board are provided in Fig. 2. After a board of the required size has been cut out using a hacksaw the two 6BA clearance mounting holes should be drilled and the single break in the copper strips should be made before soldering the components into place.

Be careful to connect the electrolytic capacitors and the two transistors the right way round. Veropins are used at points on the board where connections to off-board components will be made.

Next the component panel is mounted on the base panel of the case using 12mm 6BA bolts with 6mm spacers to hold the underside of the board clear of the metal case so that short circuits are avoided. The board is mounted so that the end of the board at which R1 and C2 are mounted is next to VR1 to VR3. The final wiring of the unit is then perfectly straightforward, and this is shown in Fig. 2.

If the unit is to be used with medium or high impedance headphones, results will almost certainly be best with the two phones wired in parallel, and this is achieved using the method of connection shown in Fig. 2 (assuming that the headphones are connected to an ordinary stereo jack plug).

With low impedance types it is better to have the phones connected in series across the output of the amplifier, and an

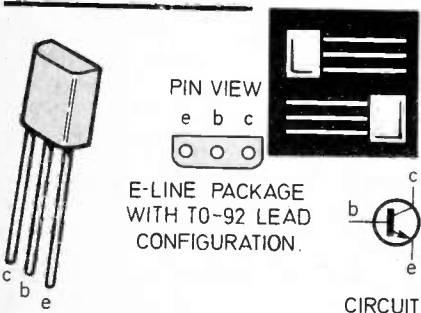
insulated stereo socket must then be used for SK2, otherwise an unwanted connection to the chassis tag of the socket will be made via the metal case and SK1. Fig. 2 shows the appropriate method of connection for this type of socket.

IN USE

The guitar lead connects to SK1, and the amplifier is automatically switched on and off when the jack plug is inserted and removed from SK1. Results will probably be best with any volume or tone control fitted to the guitar set at maximum, with the volume and tone then adjusted using the controls on the amplifier.

However, if desired, any tone controls fitted to the guitar can be used in addition to those on the amplifier to give greater control over the tone. □

2 FREE TRANSISTORS



The 2N4123 and BCY65EP are both npn silicon planar general purpose transistors. The BCY65EP is particularly suited for low noise a.f. amplifier input and driver stages and both types have been designed for use in general purpose amplifiers, oscillators, and switching applications. They are identically cased in Ferranti E-Line plastic packages with T0-92 lead-out.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	SYMBOL	BCY65EP	2N4123	UNIT
Collector-Base voltage	V_{CB}	60	40	V
Collector-Emitter voltage	V_{CE}	60	30	V
Emitter-Base voltage	V_{EB}	7	5	V
Continuous Collector current	I_C	100	200	mA
Power dissipation @ $T_{amb} = 25^\circ\text{C}$	P_{TOT}	400	500	mW
Power dissipation @ $T_{case} = 25^\circ\text{C}$	P_{TOT}	—	1.2	W
Power dissipation @ $T_{case} = 45^\circ\text{C}$	P_{TOT}	1	—	W

ELECTRICAL CHARACTERISTICS

Transition frequency (min)	f_T	125	250	MHz
Static gain* ($V_{CE} = 1\text{V}, I_C = 10\text{mA}$)	h_{FE}	80-190	—	—
Static gain* ($V_{CE} = 1\text{V}, I_C = 2\text{mA}$)	h_{FE}	—	50-150	—
Dynamic gain* @ 1kHz ($V_{CE} = 5\text{V}, I_C = 2\text{mA}$)	h_{fe}	125-250	—	—
Dynamic gain* @ 1kHz ($V_{CE} = 1\text{V}, I_C = 2\text{mA}$)	h_{fe}	—	50-200	—

*Forward current transfer ratio.

FOR YOUR ENTERTAINMENT

BY BARRY FOX

Sounds Of Video

In just five years video recording has become a way of domestic life. In Britain there are around 20 million homes and three million now have video recorders.

We have become blasé about the extraordinary trick performed by every one of these machines. It squeezes colour TV bandwidth, over 3MHz, out of half-inch tape running at less than one inch per second. This is half the running speed of audio cassette tape.

Because we take the pictures for granted, we've now started to listen to the sound. Unfortunately it doesn't bear close listening. Although the video signals are recorded by rapidly rotating heads, which scan the tape at around 10 miles an hour, the sound still has to be recorded in conventional linear fashion, at under one inch per second.

To make matters worse the audio tracks are very narrow, and when they are split into stereo with a guard band in between, they become absurdly tiny. The audio tracks on a V2000 video recorder, for instance, are just 0.65mm wide in mono. In stereo, each half track is just 0.25mm wide, with a guard band of 0.15mm in between. Although noise reduction systems can reduce hiss, there is no getting away from the fact that the sound of video is severely handicapped.

New Approach

This is why the firms behind both the Japanese formats, Beta and VHS, have developed a completely new type of sound recording system for video. Instead of recording the sound linearly, it is f.m. encoded and interlaced with the video waveform so that it is laid down by the video heads as part of the video tracks.

To make this possible a completely new type of recording technology is used. This is depth or vertical recording. It takes advantage of a natural recording phenomenon, whereby different frequencies record at different depths in a magnetic coating, with higher frequencies nearest the surface. So although the f.m. sound carrier and f.m. video carrier are recorded at the same time, by the same heads, on the same tracks, they are physically separated because they lie in different layers of the tape coating.

The Beta and VHS hi-fi sound systems won't be available for a year or so. When they do come on the market they will be compatible with existing VHS and Beta machines, because the same sound will also be recorded linearly, conventionally.

Calculating Time

Last year I wrote a short item bewailing the fact that there isn't a calculator on the market that easily computes time. You would use such a calculator to tot up the total length of several different music or speech recording segments, and then subtract them from a total available time, to see how much is left over. It would obviously be a boon for broadcast and tape editors.

After publication I immediately received a flood of letters from readers who very helpfully told me that it was possible to buy reasonably priced calculators which add time without special programming. Most are budget scientifics made by Casio. My thanks to everyone who wrote in pointing this out, but I still feel there would be a market for a calculator that did the trick more easily.

To add time with a scientific calculator you have to enter hours, minutes and seconds by keying in time as degrees and then using the sexagesimal instruction key.

Record Spoiler

You'll probably have seen reports in the popular press that CBS has developed a spoiler system to prevent unauthorised taping of records. Actually it's *not* a spoiler system and CBS is the first to admit it. But it could, in theory, prevent unauthorised taping.

A so-called spoiler signal is supposed to stop people taping records, by interfering with some vital circuit in the tape-recorder when illegal copying takes place. But if the record is to sound all right for normal listening, which is, of course, legal, the spoiler must be of too high or too low a pitch to be heard. As such it is all too easily filtered off, either deliberately or by budget equipment.

The CBS system doesn't add anything to the music signal. Instead it takes something away. A very narrow notch is filtered out of the music signal, in the mid-frequency range of around 3 or 4kHz.

This notch isn't normally audible. It's like a crack in the piano keyboard. But if a tape-recorder has a sensor circuit which can detect the notch, and switch off some vital circuit like the bias, then it will refuse to record a notched disc.

The display shows the times in decimal equivalents.

You then do the sum and convert decimals back into hours, minutes and seconds by using the inversion and sexagesimal keys. The sequence works perfectly well, but it's hardly a simple, everyday approach to totting up time!

My point was, and still remains, that the calculator manufacturers seem to be missing out on a useful feature that could be easily offered, simply by relabelling and dedicating a few keys, and combining the unit with a clock and stopwatch timer. We've already got calculators with clocks and stopwatch timers, and we have already got scientifics with sexagesimal arithmetic.

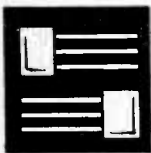
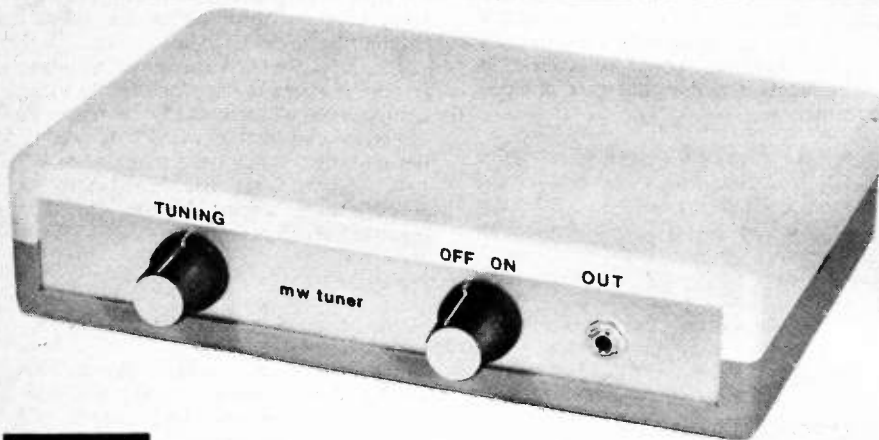
So why not package these features together for people who are more interested in adding time, than making science calculations. It's always been my sneaking suspicion that only a relatively few people who buy scientific calculators ever fully understand their functions, let alone exploit them.

Although CBS has proved that the system can work well, it relies on legislation to make tape-recorder manufacturers build in the necessary sensor and switch circuits. Realistically it's a non-starter, because there are now so many recorders without switches and discs without notches that it's too late to make the system a legal requirement.

The real irony is that the idea of marking music with an identifying notch, rather than adding a trigger signal, is at least ten years old. In the USA, radio pioneer, Murray Crosby patented something very similar for keeping tabs on the number of times radio commercials are broadcast. Other radio stations, like the British Forces Broadcasting Service, use notched trigger signals to control its tape recorders.

The record industry knew all about these notch systems back in the mid '70s but did nothing about using them to curb home taping. But, of course, at that time record sales were booming, and no-one bothered about home taping. Now, with the recession and interest in video cutting into record sales, the record industry is looking for someone or something to blame.

MW PERSONAL RADIO



A SIMPLE PROJECT FOR YOUR FREE TRANSISTORS

BCY65EP—2N4123

ALTHOUGH based on just two transistors this simple design provides good reception over the medium-wave broadcast band and gives more than adequate volume from any station that is received at reasonable strength.

The output is suitable to drive a crystal earphone, high impedance headphones, or virtually any amplifier. The unit has also been tried with a pair of medium impedance hi fi headphones and these gave excellent results.

Apart from the use of headphones, an earphone, or whatever, the unit is completely self-contained as it has an internal ferrite aerial and does not need any additional (external) aerial to give good results. Power is provided by a 9-volt battery, and many hours of use are available from each battery as the current consumption of the circuit is only about 3.5mA.

The unit is very easy to construct and it should be well within the capabilities of a beginner if reasonable care is taken during construction.

CIRCUIT DESCRIPTION

Most radio receivers use a superheterodyne (*superhet*) circuit, but while circuits of this type admittedly give excellent performance, they are relatively complex and expensive to build, and they can also be awkward to set up ready for use once completed. In this design the much more simple tuned radio frequency (r.f.) type of circuit has been used, and although this does not give results that are comparable to a *superhet* circuit, a perfectly adequate level of performance

can be achieved. Fig. 1 shows the circuit diagram of the set.

TR1 and TR2 are used in a two-stage direct coupled amplifier, and both devices are used in the common emitter mode so that the circuit as a whole has a very high level of gain. The circuit is biased so that there is approximately 1.45 volts at the collector of TR1 and about 0.7 volts at the emitter of TR2, and negative feedback through bias resistor R2 helps to stabilise these bias levels.

For example, if the collector of TR1 should go more positive for some reason, the emitter of TR2 would go positive by a similar amount, a higher base current would be fed to TR1 via R2, and TR1's collector voltage would therefore be

reduced. The collector and emitter currents of TR2 are virtually identical, and as collector resistor R3 is about six times higher in value than emitter resistor R4, the voltage developed across R3 is about six times higher than that present across R4.

With 0.7 volt developed across R4 this gives about 4.2 volts across R3, and the output of the amplifier is therefore biased to a potential that enables a high peak-to-peak output voltage to be obtained without clipping and serious distortion of the signal.

C7 is used to decouple R4 so that the negative feedback that this resistor would otherwise introduce is eliminated and TR2 is able to provide a high level of voltage gain. With only two stages of amplification in the circuit it is essential that they both provide high gain if acceptable results are to be obtained.

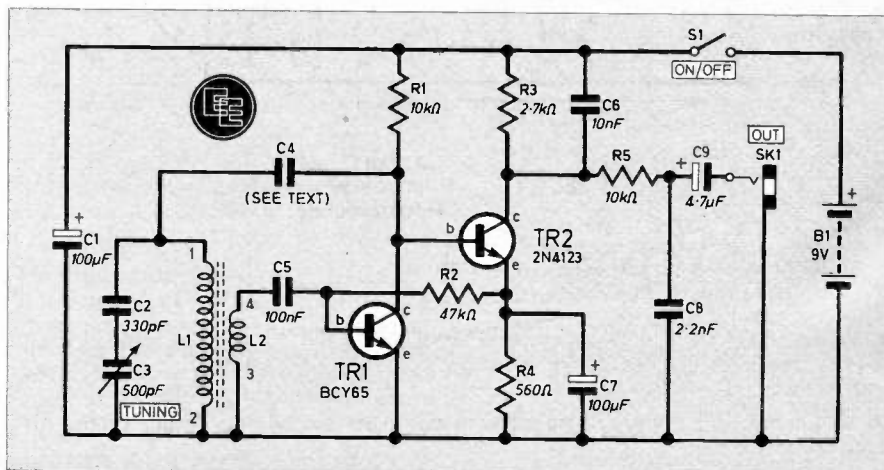
FERRITE AERIAL

The input signal for the amplifier is provided by ferrite aerial L1/2, and C3 is the tuning capacitor for the latter. The component used in the C3 position was chosen because it is small and inexpensive, but it has a somewhat higher maximum capacitance than is really needed. C2 is therefore used in series with C3 to effectively reduce its maximum value to a suitable figure. The output of the aerial is coupled to the input of the amplifier via a low impedance coupling winding (which ensures an efficient signal transfer) and d.c. blocking capacitor C5.

Filtering (r.f.) is provided at the output of the amplifier by C6, R5 and C8, and the remaining audio frequency signal is coupled to output socket SK1 by C9. Of course, r.f. filtering alone is not sufficient to produce a.m. demodulation, and rectification of the signal is also needed. This rectification is actually an inherent feature of the amplifier, and it occurs due to the variation in gain that occurs with changes in the collector current of a transistor.

On one set of half-cycles the current through TR2 increases and produces a

Fig. 1. Complete circuit diagram of the MW Personal Radio.



consequent increase in gain, while on the other set of half-cycles the collector current and gain of TR2 both reduce. As TR2 is handling a fairly high signal level the gain difference on the two sets of half-cycles is quite large, and reasonably efficient rectification and detection is produced.

STATION SELECTION

Tuned radio frequency receivers have two main deficiencies; a lack of sensitivity and a lack of selectivity. A receiver which has good selectivity can pick-up just a small part of the frequency range covered by the set so that it can pick out just one station even though the band may contain numerous strong stations with crowding in places. A simple t.r.f. circuit such as this one has all the selectivity provided by a single-tuned circuit (ferrite aerial L1/2 in this case), and it is this that gives mediocre results in this respect.

A simple way of improving both sensitivity and selectivity is to use positive feedback, or regeneration as it is often termed in this context. This simply entails coupling some of the radio frequency signals from the output of the amplifier and feeding it back into the ferrite aerial so that a boost in gain is provided.

There will be more feedback (and thus a greater boost in gain) near the centre of

the receiver's passband, and this effect produces the improved selectivity.

In this circuit the feedback is from the collector of TR1 to the ferrite aerial via capacitor C4. The latter has a very low value, and is in fact made from a couple of small pieces of insulated wire. The circuit is arranged so that the feedback is of the positive type so that the feedback signal is added to the input signal and a boost in gain and selectivity are obtained. Negative feedback would be useless in this application and would give reduced performance.



COMPONENT BOARD

A 0.1 inch matrix stripboard panel measuring 12 copper strips by 23 holes accommodates most of the components. Details of this board and the wiring of the receiver are illustrated in Fig. 2. There are no breaks in any of the copper strips.

Construction of the board starts by cutting out a board of the specified size using a hacksaw, and any rough edges are then smoothed using a file. Next the two 6BA clearance holes are drilled and the components are soldered into place leaving TR1 and TR2 until last. Be careful to fit C1, C9, C7, TR1 and TR2 the right way round, and make sure that none of the copper strips are shorted together by solder splashes.

It is helpful to fit Veropins to the board at points where connections to off-board components will eventually be made as the board can then be wired to the rest of the unit after it has been mounted in the case. Alternatively the board can be wired into the circuit and then bolted in place in the cabinet.

CASE

The case used for the prototype is a Verobox having approximate outside dimensions of 180 x 120 x 40mm. This gives a smart appearance and makes construction of the unit very simple and straightforward, but any case of largely non-metallic construction and about the same dimensions should be suitable, and with a little ingenuity it should be possible to use a much smaller case if desired.

Note that it is important that the case is largely of non-metallic construction

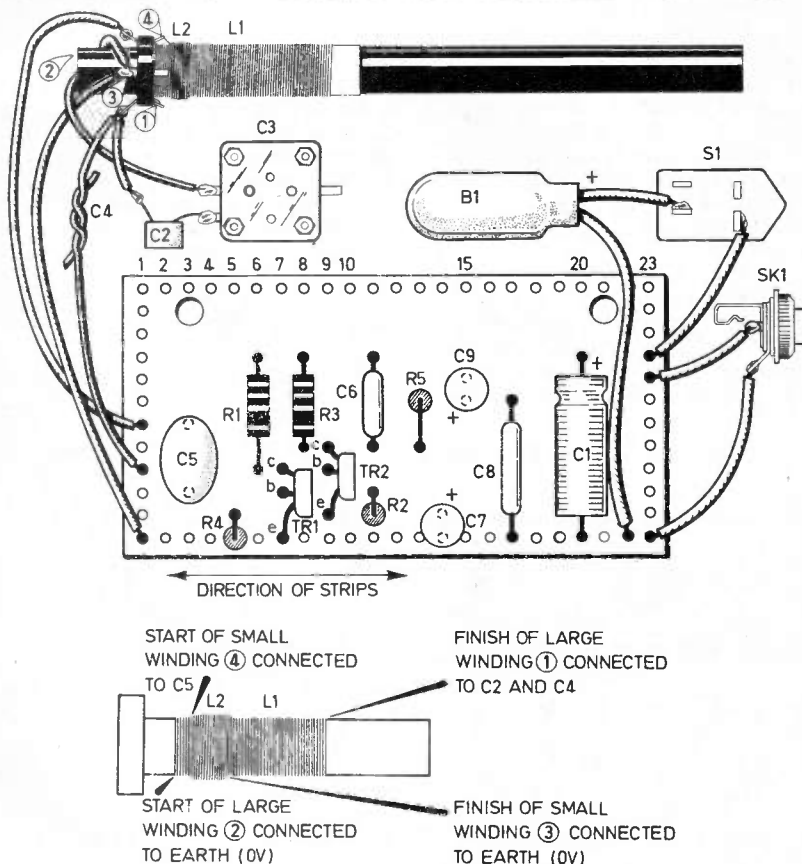


Fig. 2. Layout of component board with complete interwiring details of the off-board components.

COMPONENTS

Resistors

R1,5	10k Ω (2 off)
R2	47k Ω
R3	2.7k Ω
R4	560 Ω
All	$\frac{1}{4}$ W carbon $\pm 5\%$

See
**Shop
Talk**
page 281

Capacitors

C1	100 μ F axial elect.
C2	330pF ceramic plate
C3	500pF variable
C4	see text
C5	100nF mylar
C6	10nF mylar
C7	100 μ F 10V radial elect.
C8	2.2nF mylar
C9	4.7 μ F 63V radial elect.

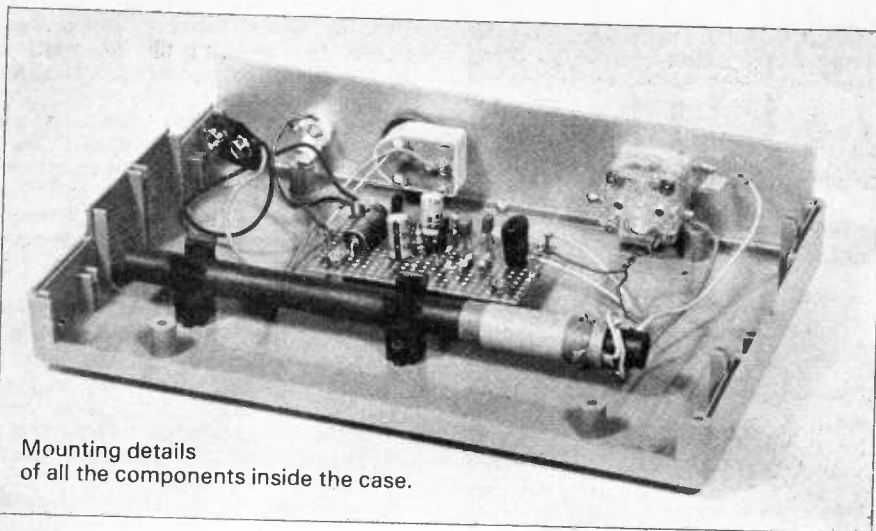
Semiconductors

TR1	BCY65EP npn silicon
TR2	2N4123 npn silicon

Miscellaneous

L1	medium-wave ferrite aerial (Denco 5FR)
S1	rotary on/off switch
B1	9 volt (PP3 or PP6)
SK1	3.5mm jack socket
Verobox	75-3007C size 180 x 120 x 40mm; 0.1 inch matrix stripboard, 23 holes by 12 strips; two control knobs; battery connector; 6BA fixings.

Approx. cost
Guidance only **£7.00**



Mounting details of all the components inside the case.

since metal would have the effect of screening the ferrite aerial so that it would be unable to pick-up any significant signal, and the unit would obviously fail to work at all. Although the Verocase used for the prototype does have aluminium front and rear panels, these are quite small and do not have any significant screening effect on the aerial.

The two controls and the output socket are mounted on the front panel. The component specified for C3 has provision for screw mounting, but it may be easier to glue it in place using a good quality general purpose adhesive.

The ferrite aerial used in the prototype is an Ambit aerial coil type MWC2 mounted on a 140 x 9.5mm ferrite rod type FRA, which is in turn mounted on the base panel of the case towards the rear of the unit using two 6.35mm 6BA bolts and two mounting clips type FRPC. However, the circuit has been found to work well using a Denco type MW5FR aerial which is supplied complete with a 140 x 9.5mm ferrite rod. This aerial can be mounted using a couple of large P-grips.

The circuit should work using a normal medium-wave ferrite aerial having a

small coupling winding. It is essential for the aerial to be connected correctly, and reference to Fig. 3 plus a close inspection of the aerial coil should clarify matters.

Once all the components have been mounted in the case the final wiring of the unit can be completed. This includes the fitting of C2 which is mounted direct on C3, as shown in Fig. 2. Capacitor C4 just consists of two pieces of insulated wire about 75mm long and connected to the appropriate points in the circuit.

Single-strand wire is better than the multi-strand type since single-strand wire will not tend to spring apart and unwind itself. Initially these two wires should not be twisted together and should just be positioned close to one another.

Picture showing the front panel of the finished article.



TUNING ADJUSTMENT

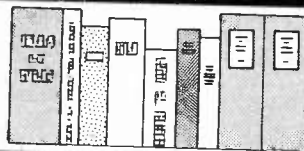
To an extent the coverage of the receiver is controlled by the position of the aerial coil on the ferrite rod. The aerial coil can be left at any position on the ferrite rod that provides full coverage of the medium-wave band, and its exact position should not be very critical since C3 gives slightly more than complete coverage of the band.

In order to obtain good results from the circuit it is essential to have C4 set correctly. With the two wires forming C4 a little way apart it will probably be possible to receive a few stations at moderate volume, but the set will probably lack reasonable selectivity. Twisting the two wires together should improve both sensitivity and selectivity, but if they are twisted together over too great a length oscillation will occur at certain settings of C3.

This will be heard as a tone of varying pitch as the set is tuned through a station, and proper reception is not possible with oscillation present. The two wires forming C4 should be twisted together as much as possible without oscillation being evident at any setting of C3. The receiver will then have optimum performance and should provide good reception.

Remember that the ferrite aerial is directional, and if necessary the unit can be turned to peak the wanted station and to null any interfering transmission. It is possible that slight overloading could occur in strong reception areas, and the set can then be turned to reduce the strength of the received signal to a satisfactory level. □

BOOK REVIEWS



AN INTRODUCTION TO VIDEO

Author D. K. Matthewson
Price £1.95 Limp edition
Size 180 x 110mm 87 pages
Publisher Bernard Babani Ltd
ISBN 0-85934-075-9

FOR those who are relative newcomers to home video this book will provide some very useful information and answers. In some cases it could possibly save you money.

The book deals with the circuitry and mechanical features of typical recorders with Beta and VHS format. The book contains helpful hints on tape editing, copying and the use of video cameras.

R.A.H.

THE ART OF PROGRAMMING THE ZX SPECTRUM

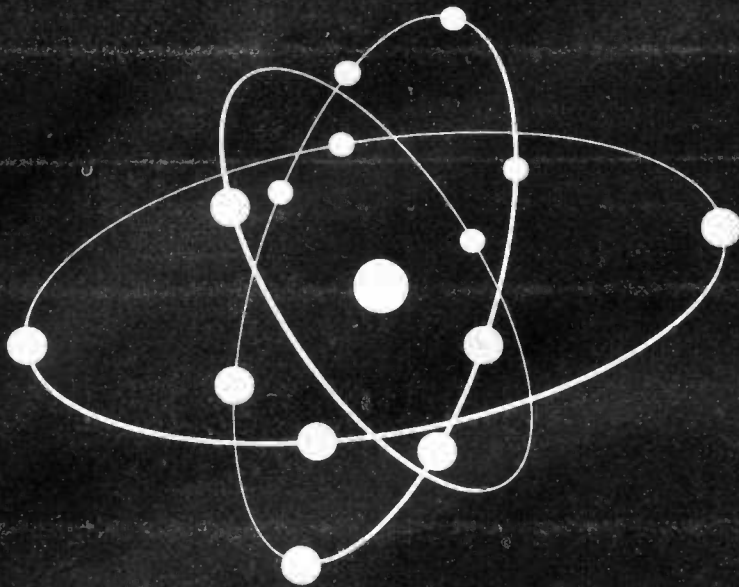
Author M. James
Price £2.50 limp edition
Size 180 x 110mm. 138 pages
Publisher Bernard Babani Ltd
ISBN 0-85934-094-5

THIS book provides a very detailed and comprehensive guide to programming the Sinclair ZX Spectrum. Each area of programming is explained in easy language and each example program is accompanied by a program listing.

Not only is this book useful for an experienced programmer but it will also provide the beginner with a useful introduction to computing.

R.A.H.

ELECTRONICS and the ELECTRON



Part 1 INTRODUCING THE ELECTRON

BY J. B. DANCE

WHAT is this peculiar particle, the electron, from which our hobby of electronics has derived its name? An electron is an extremely minute object weighing less than one thousand million million million millionth of a gramme; such an object is quite beyond our imagination. To add to our confusion an electron may behave as a particle or as a wave, depending on the particular experiment one performs!

THE UBIQUITOUS ELECTRON

Electrons are found in enormous numbers in all types of matter. When chemical changes occur (such as the burning of coal or the use of food in our bodies), it is the electrons in the matter undergoing the change which alter their orbital positions and so produce new chemicals.

If one combs one's hair or rubs a piece of plastic to generate electric charges, one merely removes an extremely minute fraction of the electrons present in a material and transfers them to another material, thus creating very high voltages.

Electrons are therefore, to all intents and purposes, responsible for almost all of the phenomena of everyday life. Indeed, it is only the electrostatic repulsion between the electrons of our bodies and those of other materials, which prevents us from sinking rapidly into the ground towards the centre of the earth under the force of gravity, since all matter is almost entirely empty space (except possibly the extremely dense matter of neutron stars).

The compressed air inside a vehicle tyre supports the weight of the vehicle as a result of the bombardment of huge numbers of air molecules on the inside of the tyre. However, if we look at the molecules of air more closely, we find it is the repulsion between the electrons of the air and of the car tyre which supports the vehicle. Similarly the braking force of a car is essentially provided by inter-actions between the electrons of the tyre and electrons of the road surface.

ELECTRONICS

However, in this article we shall be mainly concerned with the flow of elec-

trons in various materials and in electronic tubes, since it is this flow which is an electric current and which makes our hobby of electronics possible. Initially electronics was a term applied to the use of vacuum tubes (valves) in which electrons flowed. In the early years of this century, electronics was a term exclusively reserved for what we now call nuclear instrumentation (that is, the use of electronic devices for nuclear particle detection and measurement).

The term electronics subsequently underwent some change in use with the rapid developments in radio communication. In the 1950s and 1960s the term electronics gradually became the science of using semiconductor materials and devices rather than vacuum tubes. However, one still meets the vacuum tube in the cathode ray tubes of the television receiver, in certain microwave electronic tubes and in a few other applications where no semiconductor device has yet been found to replace the electronic tube both for performance and economy.

The flow of electrons is responsible for all aspects of electronics. It would seem equally reasonable to call the whole of chemistry (or perhaps even the whole of science) "electronics". However, the name "chemistry" and the names of most of the other sciences had become firmly established even before the electron had been discovered.

HISTORY

The discovery of the electron was one of the most important of a number of discoveries made near the turn of the century which completely revolutionised our fundamental ideas of physics. In 1895 Röntgen discovered X-rays using discharge tubes and in the following year Becquerel found radioactivity using photographic techniques.

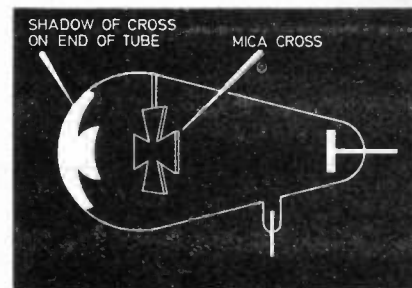


Fig. 1.1. A typical Maltese Cross tube for demonstrating the emission of cathode rays.

In 1897 J. J. Thomson discovered the electron and this greatly assisted our understanding of the constitution of matter. Important discoveries in the nuclear field soon followed together with Einstein's *Special Theory of Relativity*.

It was known that if a high potential was applied to an evacuated tube, such as that shown in Fig. 1.1, the glass of the tube glowed with a green light. If an ob-

ject such as a mica cross is placed in the position shown, a shadow of this cross is thrown onto the end of the tube. This is known as the Maltese Cross Experiment.

It seemed that something was being emitted from the negative electrode (the cathode) and was travelling more or less in straight lines towards the glass at the end of the tube. What was this "something"? Initially it was assumed that some form of ray was given out by the cathode and these rays were therefore called "cathode rays"—a name which still persists in our cathode ray tubes.

German scientists of that time believed that the rays were some form of invisible light, whereas English scientists felt that they were minute particles of negative electricity. The latter proved correct.

It was found that cathode rays could penetrate a thin sheet of aluminium. If allowed to fall onto a light paddle wheel made of mica, they would cause the wheel to rotate on its axis. In 1895 J. Perrin collected cathode rays in a cylinder and used an electroscope to show they have a negative charge.

We now believe that cathode rays consist of a beam of electrons. Apparently the name "electron" was first suggested by Stoney in 1894 for the particles of cathode rays, but it was not clear at that time whether the rays consisted of a continuous flow of some material or whether individual particles were really present.

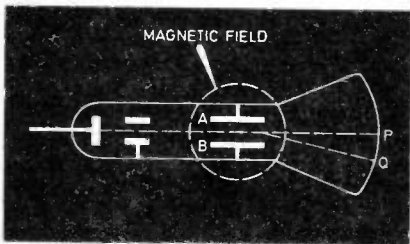


Fig. 1.2. The vacuum tube used by J. J. Thomson.

THOMSON'S EXPERIMENT

In 1897, J. J. Thomson measured the ratio of the charge of the particles to their mass and this enabled scientists to conclude that individual electrons were present in the cathode rays. The type of tube used by Thomson is shown in Fig. 1.2.

A narrow beam of cathode rays is produced by an arrangement known as an electron gun. The beam passes through a region in which either an electric field or a magnetic field or both can be applied to the rays, the direction of the two fields being perpendicular to one another.

The electric field is induced by applying a voltage to the electrodes marked A and B, whilst the magnetic field is produced by a coil outside the tube in the region indicated by the dotted circle of Fig. 1.2.

Thomson applied a magnetic field which caused the electrons to be deflected from P to Q. The position where the beam met the end of the tube was made

visible as a green glow, as in the modern cathode ray tube. Thomson then applied an electric field between the plates which was just sufficient to return the beam to its former position at P.

The magnetic and electric fields were now producing equal and opposite effects. Thomson now entered values for the field strengths into suitable equations in order to obtain a value for the ratio of the charge/mass for the particles of the beam. He was also able to show that the velocity of the particles was about 10 per cent of that of light.

The important discovery made by Thomson was that the charge/mass ratio of cathode rays was always the same no matter what material was used for the cathode. This strongly implied that all electrons are alike and that they are universal constituents of all matter.

The charge of an electron (given the symbol e) was measured by J. J. Thomson and by Townsend independently in 1897-98, but Milikan developed a more accurate technique during the period 1909-1916. This charge was found to be the same as that carried by the hydrogen ion (or any other monovalent ion) in solution and is now taken as the fundamental unit of electric charge. No particle has yet been found which has a charge which is not equal to e or some whole number multiple of e .

As the charge/mass ratio, e/m and the charge e were both known, the mass of the electron, m , could be found. Electrons are one of the lightest known particles, having a mass of only about $1/1840$ of that of the hydrogen atom, this being the lightest known atom.

Sir William Bragg has said that Sir J. J. Thomson (1856-1940) "more than any other man was responsible for the fundamental change in outlook which distinguished the physics of this century from that of the last." He led a team at the Cavendish Laboratory, Cambridge, where he was professor of experimental physics. He was succeeded by his pupil, Rutherford.

THE CATHODE RAY TUBE

Thomson's tube of Fig. 1.2 was the forerunner of the modern cathode ray tube. A typical c.r.t. is shown in Fig. 1.3.

Hot objects emit electrons and

therefore a heater is used at the end of the tube. This has the advantage that a much lower voltage is adequate to operate the tube than in heater-less tubes in which the electrons had to be torn out of the cathode by the high electric field produced by a high operating voltage. In addition, the use of a relatively low operating voltage, not exceeding a few kilovolts, removes the danger of X-rays being formed. When cathode rays (that is, electrons) strike a surface, they lose their velocity and X-rays are formed.

The grid or modulator electrode controls the number of electrons which pass along the tube through a hole in the anode. They then pass through two sets of deflector plates; one set of plates can deflect them vertically and the other set horizontally. Finally, they strike the fluorescent screen where a part of their energy is converted into light.

The cathode ray tubes used in television receivers are somewhat similar to the oscilloscope type of tube shown in Fig. 1.3, but the electron beam is deflected by means of a magnetic field generated by coils placed around the neck of the tube.

A FUNDAMENTAL PARTICLE

The electron is one of the fundamental particles of nature. The other two fundamental particles found in atoms are the proton and the neutron. These last two particles are found in nuclei, a typical nucleus being about 10^{-13} cm in diameter.

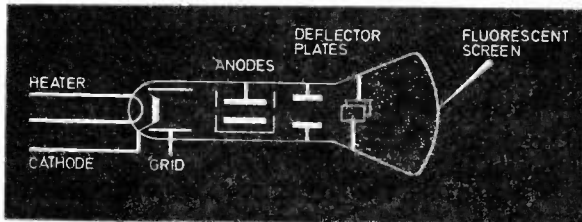


Fig. 1.3. A typical modern cathode ray tube as used in oscilloscopes.

The electrons are said to be in orbits around nuclei, but whether they actually circle around the nucleus according to Bohr's theory will not be discussed here.

The particles emitted in a type of radioactive decay known as beta decay are electrons. Beta particles are identical to electrons. Many other types of nuclear particles are known, such as neutrinos, muons and kaons, but they need not concern us.

Next month: Conduction and how electrons flow in various materials.

ELECTRON DATA

e = negative charge of an electron = $1.6021892 \times 10^{-19}$ coulomb.

m = mass of the electron = 9.109534×10^{-28} gramme.

(The mass of the electron can also be expressed in terms of energy, using Einstein's equivalence of mass and energy; it is equal to 0.5110034 MeV where 1 MeV = 1 million electron volts.)

As far as is known, the electron is stable indefinitely; it has been shown to have a half-life of over 10^{21} years.

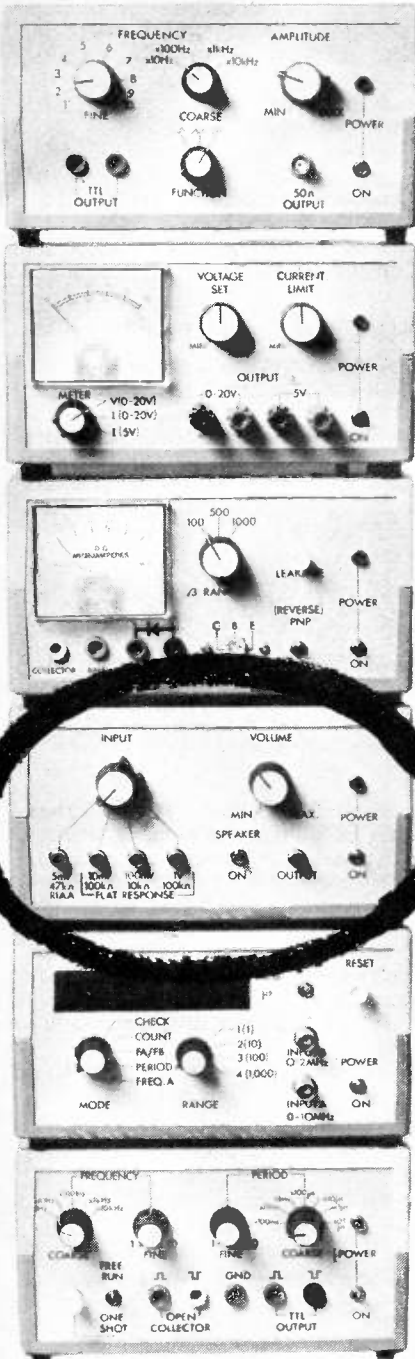
TEST GEAR 83 LABORATORY AMPLIFIER

UNIT THREE

BY J. R. W. BARNES

THE TEST GEAR 83 SERIES CONSISTS OF:

DUAL POWER SUPPLY ● FUNCTION GENERATOR ● TRANSISTOR TESTER
PULSE GENERATOR ● LABORATORY AMPLIFIER ● FREQUENCY METER



PROJECTS which involve sounds of one kind or another are both popular and common among hobbyists. They range from radios and musical instruments to alarms and special effects generators. In all cases some form of amplifier and loudspeaker must be used to convert the electrical signals to sound waves.

The unit described here is a small half watt amplifier, with a pre-amplifier section designed to accept a very wide range of inputs. It has a small internal speaker which produces acceptable results, but better fidelity is obtained with an external speaker.

In line with the other units in the series the amplifier is mains powered and housed in a matching case.

FOUR INPUTS

The amplifier has four inputs. Three of these have a flat response, that is, provide the same degree of amplification across the audio band. The fourth input has RIAA magnetic pick-up equalisation.

A few words of explanation may be needed here. When records are recorded, the bass end of the spectrum is attenuated and the high frequencies boosted. There are many reasons for this, the most important being to improve the signal-to-noise ratio and to prevent the stylus bouncing out of the tracks with heavy bass.

To obtain the correct sound on playback, the frequency response of the amplifier must boost the bass and reduce the treble. To ensure compatibility, a standard frequency response has been adopted called RIAA. This facility will enable record decks with magnetic cartridges to be tested without the need for a complete hi fi in the workshop.

The Laboratory Amplifier is only single channel (mono) but two could be used for stereo. The instrument case is not an ideal loudspeaker enclosure and better reproduction is obtained using an external speaker. Tone controls were

thought to be unnecessary and as a consequence have been omitted.

The amplifier has a very wide bandwidth, the high frequency breakpoint being at around 200kHz, whilst this does not conform to the normal practice of audio amplifier design, it allows the unit to be used with confidence as a pre-amplifier to an oscilloscope or an a.c. voltmeter.

CIRCUIT DESCRIPTION

The circuit is designed around a monolithic audio amplifier integrated circuit, the LM386 (IC2). Intended for a wide range of consumer products including radios, tape recorders and televisions, this i.c. was designed to operate from a single supply and with minimum external components. These features make it ideal for this application. The circuit diagram is given in Fig. 1.

The amplifier gain is set internally to 20 (26dB). For those unfamiliar with the system of measuring gain in decibels remember that multiplication of two numbers can be carried out by adding the logarithm of each number and taking the antilog of the result. Decibels are conveniently a logarithmic scale, so if the gain of two stages of an amplifier are known in decibels, the combined gain is simply the sum. The gain in dB is obtained by:

$$\text{Gain in decibels} = 20 \times \log_{10} \text{gain}$$

For example, a gain of 2 is equivalent to 6dB. Attenuation can also be expressed in decibels, a division by a factor of 2 is equivalent to -6dB.

Using the formula:

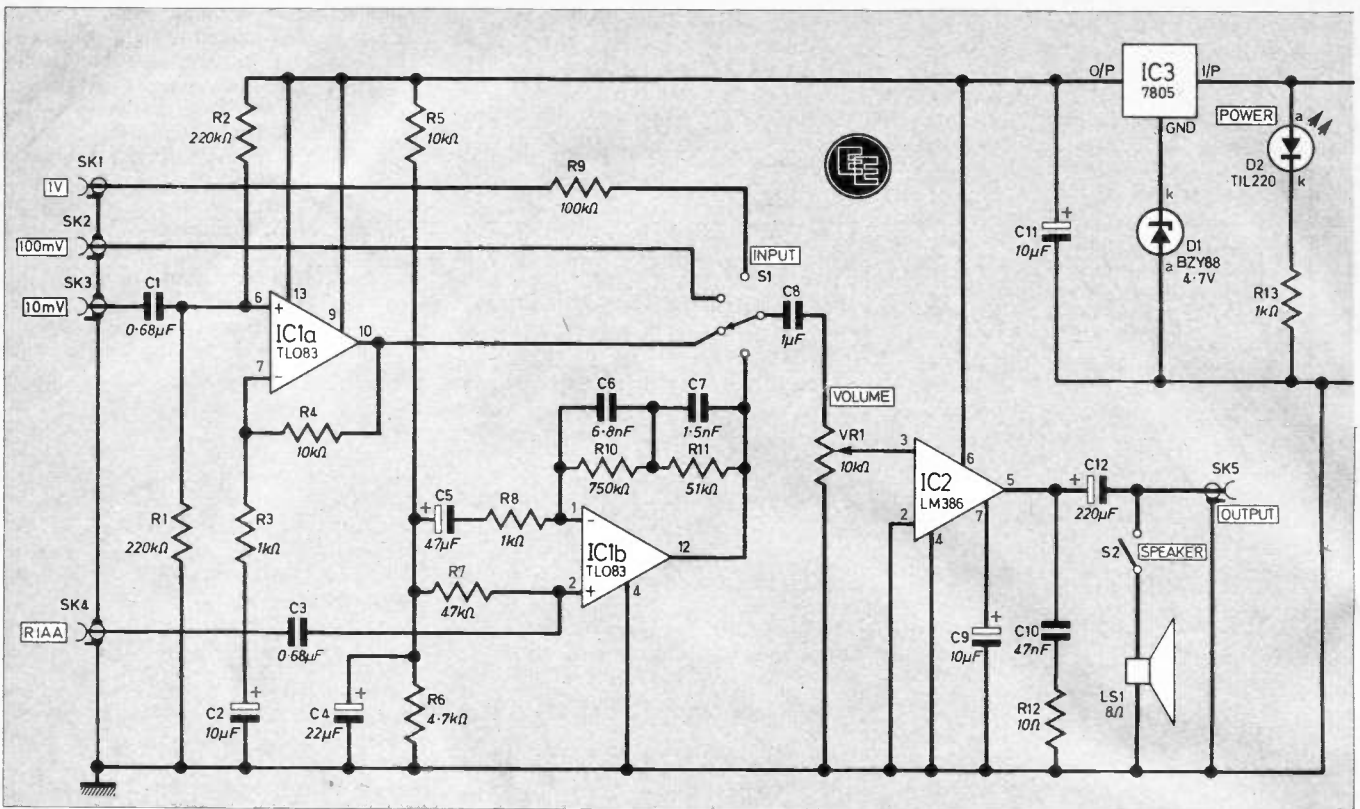
$$\text{power} = \frac{V^2}{R}$$

for the required output of half a watt with an 8 ohm loudspeaker, an output voltage of 2V r.m.s. is required. Since the amplifier has a gain of 20, the corresponding input signal (for full output) is 100mV.

SPECIFICATION

Output power: 0.5W
Input (RIAA): 5mV, 47kΩ
Inputs (flat): 10mV, 100kΩ
100mV, 10kΩ
1V, 100kΩ

Bandwidth: 20Hz-200kHz
Output: Internal speaker or via phono socket
Controls: Input select, volume, speaker on



The input selector switch S1 allows the LM386 to be driven directly from the 100mV socket. The less sensitive, 1V input is implemented in the form of a resistive attenuator formed by R7 and the volume control potentiometer VR1.

The active components for the pre-amplifier are contained in the dual op-amp, IC1. One half of this is configured as an amplifier with a gain of 10 and a flat frequency response. This is the 10mV input into IC1a. The other half, IC1b, is configured as a RIAA equalised amplifier.

This amplifier has a gain of around 36dB at 1kHz. The frequency response of the amplifier is shown in Fig. 2. Notice how the gain changes with frequency. This type of graph, showing the relationship between gain and frequency, is commonly called the transfer function.

The power supply is a mains derived 10V d.c. supply regulated by IC3, a 5V regulator and D1, a 4.7V Zener diode in the common line of IC3.

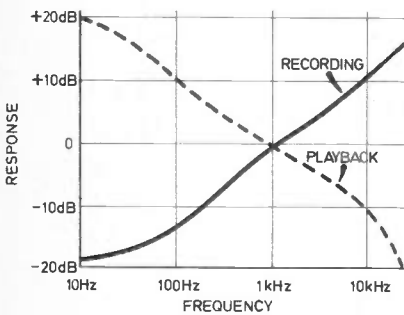


Fig. 2. The record characteristics for an RIAA equalised amplifier.

COMPONENTS

Resistors

R1,2	220kΩ (2 off)	R6	4.7kΩ	R10	750kΩ ±2%
R3,8,13	1kΩ (3 off)	R7	100kΩ	R11	51kΩ ±2%
R4,5	10kΩ (2 off)	R9	47kΩ	R12	10Ω

All 1/4W carbon ±5% unless otherwise stated

Capacitors

C1,3	0.68μF Siemens (2 off)
C2	10μF, 10V tantalum
C4	22μF, 40V elect. radial lead
C5	47μF, 10V tantalum
C6	6.8nF Siemens
C7	1.5nF Siemens
C8	1μF polyester
C9	10μF, 25V elect. radial lead
C10	47nF Siemens
C11	220μF, 16V elect. radial lead
C12	10μF, 16V elect. radial lead
C13	470μF, 25V elect. radial lead

Semiconductors

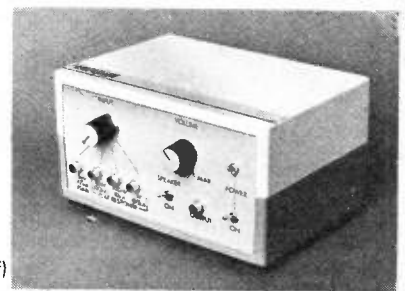
D1	BZY88 C4V7 4.7V Zener
D2	TIL220 0.2in red l.e.d.
D3-6	1N4002 silicon rectifier (4 off)
IC1	TL083 dual op-amp
IC2	LM386 op-amp
IC3	7805 5V, 1A regulator

Miscellaneous

VR1	10kΩ log potentiometer
S1	3-pole, 4-way midget rotary
S2	s.p.s.t. miniature toggle
S3	d.p.d.t. miniature mains
SK1-5	phono socket (5 off)
LS1	8Ω, 500mW loudspeaker
T1	miniature mains transformer with 12V, 500mA secondary

Single-sided p.c.b. 140 x 90mm; Verocase type 202 21036C; control knob with pointer (2 off); 7/0.2mm wire; mains lead; P-clip; l.e.d. mounting clip; 14-pin i.c. holder; 8-pin i.c. holder; Veropins; mounting hardware.

See
**Shop
Talk**
page 281



COMPONENTS
approximate
cost **£22**

CONSTRUCTION

CIRCUIT BOARD

The majority of the circuit is mounted on a single-sided printed circuit board, 140 x 90mm. The track layout is shown in Fig. 3. The prototype board was made using etch-resistant transfers and it is recommended that this method is used if constructors choose to make their own p.c.b. Veropins should be used where flying leads leave the board.

The order of construction is not

critical, but a systematic approach is best. Either work from left to right or solder in the components sequentially, working down the components list. It is suggested that holders are used for the integrated circuits as this will facilitate substitution should they become damaged through mis-use of the unit.

Before switching the unit on check the polarity of the electrolytic capacitors and diodes, the orientation of the integrated circuits, and the mains wiring for mistakes.

For safety reasons the front and rear panels should be connected to mains earth.

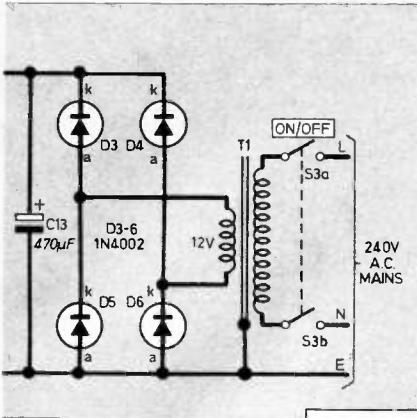
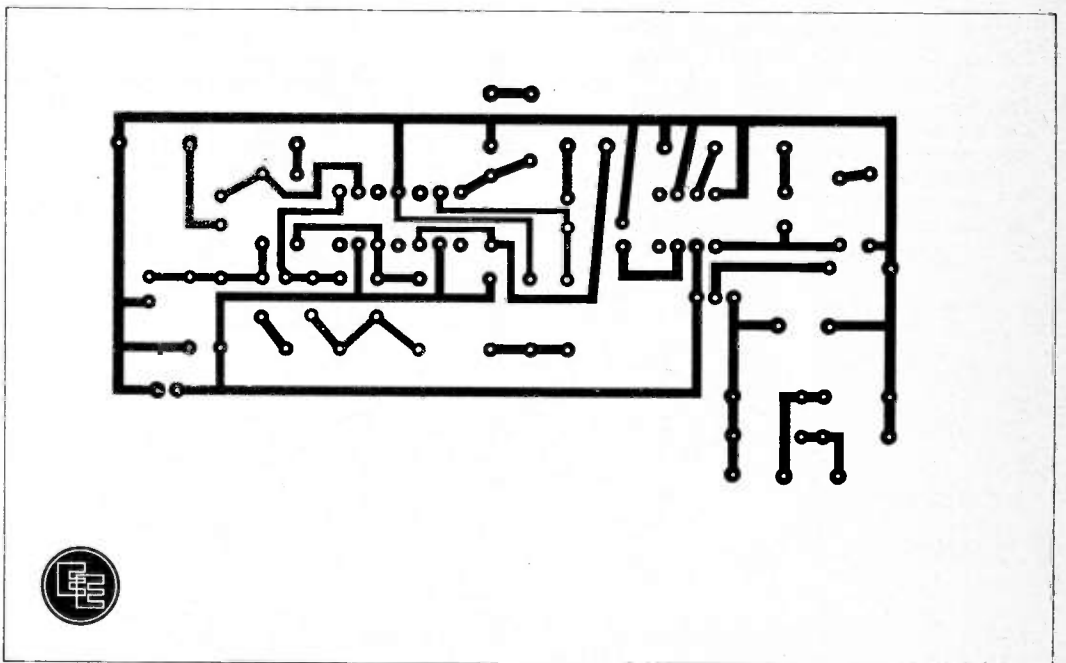
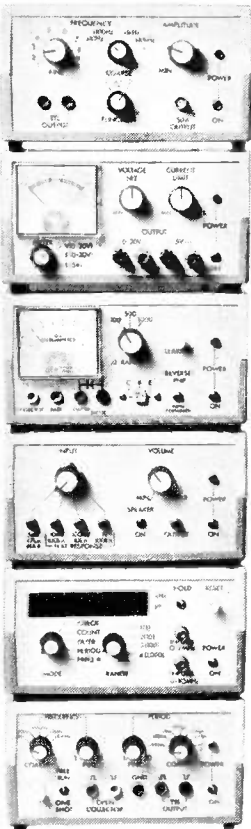
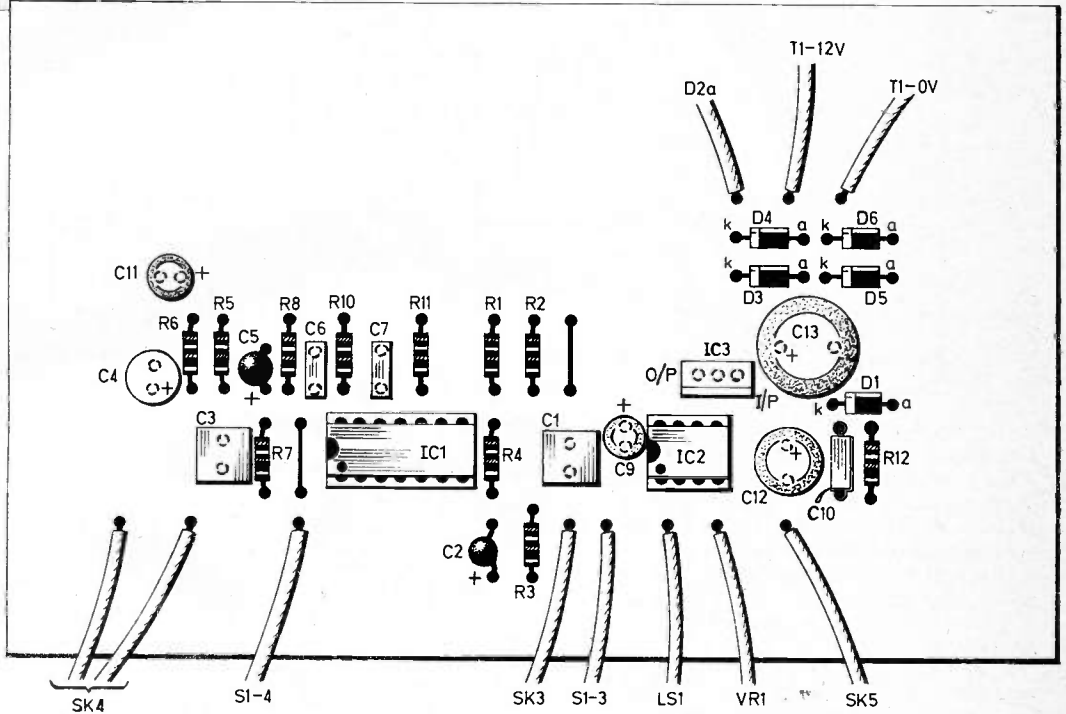


Fig. 1 (left and above). The complete circuit diagram of the Laboratory Amplifier. The labelling of the input select switch, S1, is as for the input connectors.

Fig. 3. The full size p.c.b. artwork and component layout of the Laboratory Amplifier. Four mounting holes are required and these are positioned to suit the threaded pillars in the Verobox.



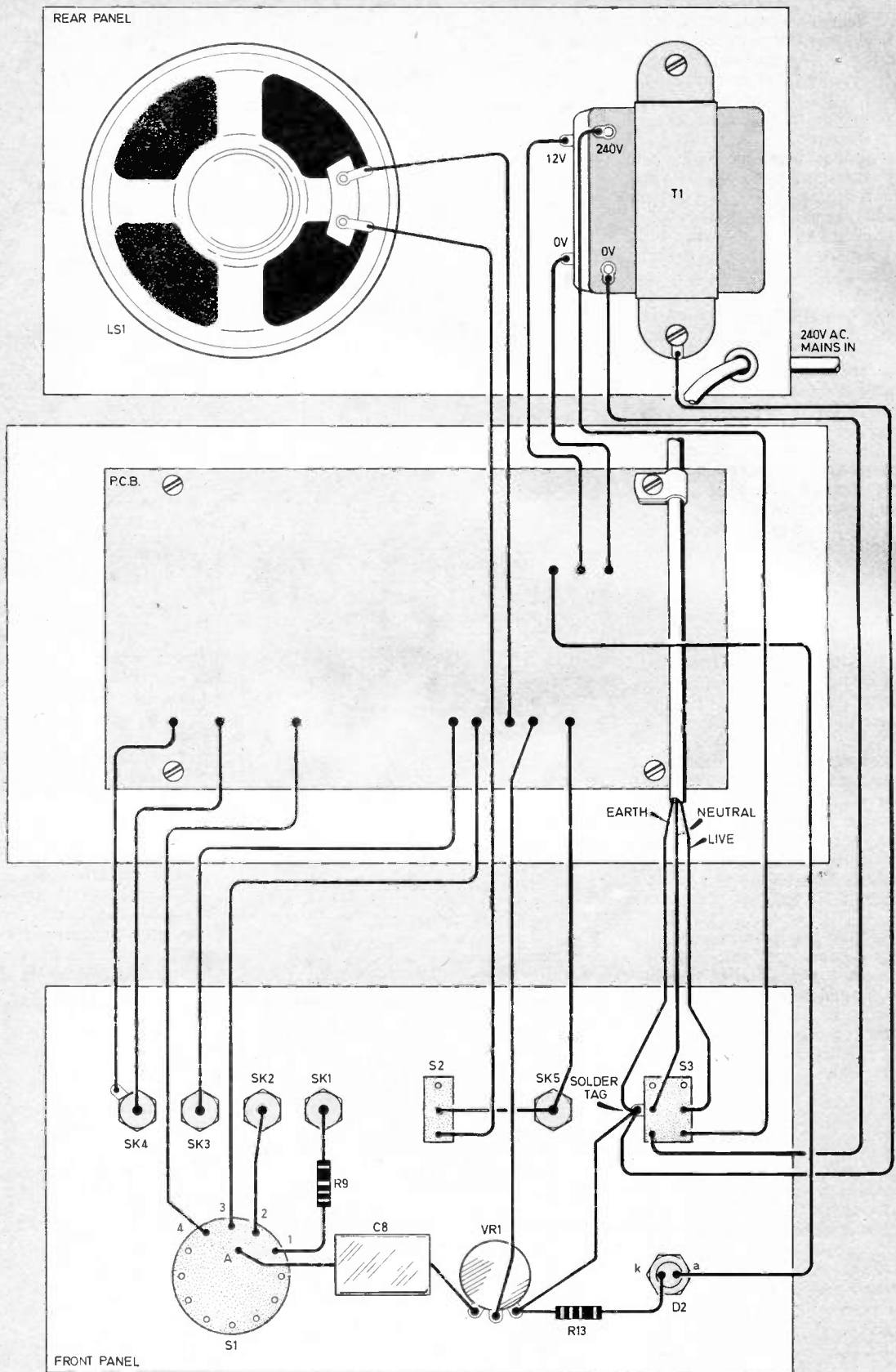


Fig. 5. Front panel layout and interwiring diagram.

CASE

The unit is built in a Verocase type 202 21036C. Begin construction with the preparation of the front and rear panels. The front panel drilling is shown in Fig. 4. On the prototype the controls were positioned to line up with the controls on other units in the series.

The rear panel must be rendered acoustically transparent to allow the sound from the loudspeaker to escape. Several options are available; a large hole covered in speaker cloth or as used on the prototype, a pattern of small holes drilled in the aluminium.

It is advisable to check the actual components before drilling any holes, because hole sizes can vary considerably.

Single hole mounting phono sockets were used on the prototype, but there is no reason why other types such as DIN style or jack sockets could not be used, depending on the constructors requirements. The l.e.d. POWER indicator used was the type enclosed in a metal housing, as these are smarter and more robust than the small plastic clips.

The final point-to-point wiring is carried out following the diagram shown in Fig. 5.

TESTING

A signal source is required for testing. Those who built the Function Generator (EE, April 1983) can use that. Constructors without a signal generator can use the earphone output of cassette player or a radio.

Connect the signal source to the 1V input socket, set the VOLUME control to about a quarter clockwise, make sure the speaker switch is in the ON position and then turn the mains on. If all is well, the sound should be heard.

Turn the signal source down and try the 100mV input. Because the other two inputs are very sensitive, a potential divider made from two resistors, should be positioned between the input and the signal source, see Fig. 6. The 10mV input should have normal balanced sound due to its flat response. However, due to the effects of the equalisation circuit, the RIAA input will have a "bassy" sound and very little treble.

IN USE

There are many uses for a small amplifier in the workshop. If you are repairing a fault on a tape deck it will enable it to be tested without the need for a full hi fi set up. The amplifier can be used as a signal tracer to isolate faulty components in audio amplifiers. The sound effects circuits which are published from time to time can be tested on a breadboard without having to build the audio amplifier section.

The unit has also been used in conjunction with an oscilloscope to examine voltages too small to be viewed directly, and it is for this reason the bandwidth has been extended to 200kHz. □

ALL DIMENSIONS IN mm

2 OFF 'A' HOLE DIA. 10
5 OFF 'B' HOLE DIA. 7.2
3 OFF 'C' HOLE DIA. 6.2

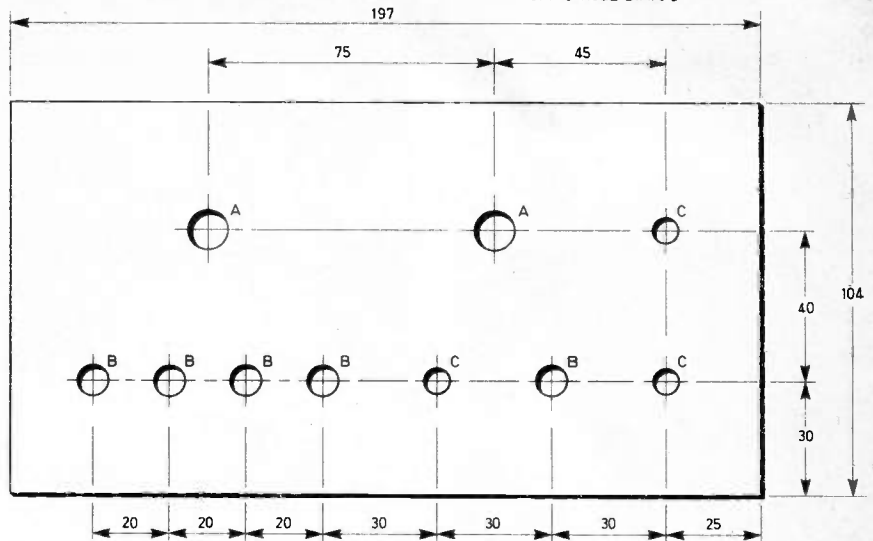


Fig. 4. Front panel drilling details.

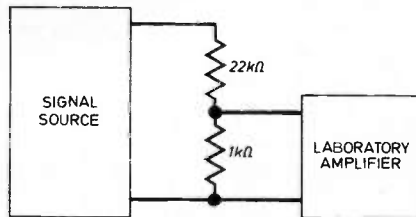
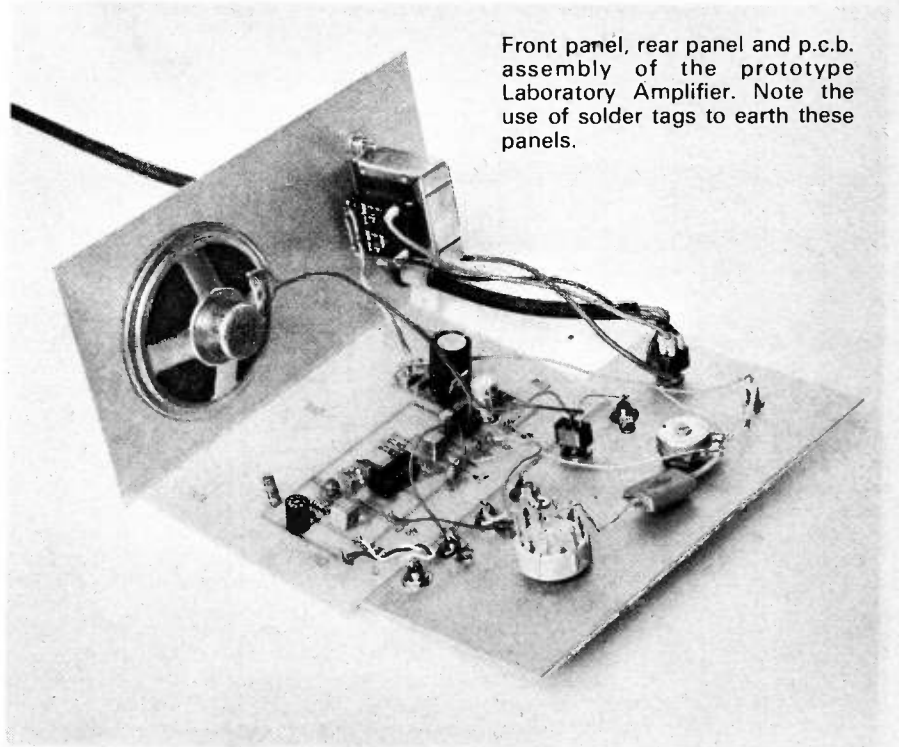
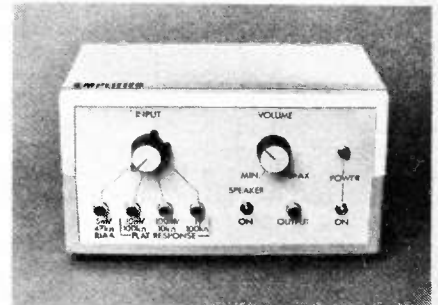


Fig. 6. Test set-up for the 10mV and RIAA inputs.



Front panel, rear panel and p.c.b. assembly of the prototype Laboratory Amplifier. Note the use of solder tags to earth these panels.

COUNTER INTELLIGENCE

By PAUL YOUNG

Behind The Counter

The term *Counter Intelligence* in the context of these articles should lead one to assume a close proximity to a sales counter, and not being associated with "Smiley's People". This is correct, but if a business is reasonably successful, the owner of that business will be pushed further and further away from the aforesaid counter. He will be called in only during those odd occasions, and sadly today they are very few.

During these last few months, having offered to help a friend of mine I have been thrown in at the deep end again, and found myself in intimate contact with the counter. It has made me realise that the lot of a shop assistant is a hard one.

My friend cannot afford more staff, and being a fairly busy shop, we are on our feet almost continuously for eight hours. Lunch consists of coffee and a sandwich, with bites taken out of the sandwich between customers. No-one is to blame, because there is no other way it could be made to work. I am lucky because I only do five days, whereas my friend invariably does seven.

Customers on the whole are friendly and understanding which eases our burden. The good ones know exactly what they want and describe it accurately or give you a clearly written list. The more inept ones will go on describing an object for five minutes after you have told them you don't stock it even though there may be several other customers waiting to be served.

It is not difficult to spot the old hand. If he asks you for six items and you have four, he will take them, while the inexperienced customer will politely tell you that he will go to a shop that stocks the lot, not realising that such a shop may not even exist.

I had an extreme example of this a few days ago when a customer asked me for three in-

tegrated circuits. I had two of them and bought them to the counter. He then asked me where he could purchase the third and I directed him to another retailer in the vicinity. To show his gratitude he said, "I will buy all three from him" and walked out. Luckily for him I am of a passive nature otherwise an incident might have occurred.

About two weeks ago, my friend took on some casual labour. A lad living on social security. He told us if it wasn't for his video recorder he would be awfully bored with life. My friend said wistfully that he wasn't even able to afford a colour television.

When asked if he would do two days the following week, he replied, "Sorry mate, I'm off to Italy tomorrow for a fortnight's holiday". My friend who hasn't been able to afford a holiday for several years looked at me and said, "I wish I knew the secret of how it's done".

Fresh Air

Ionisers having now found favour with the medical profession seem to be an acceptable accessory in the home. I built one about two years ago and though it appears to work successfully it has posed one or two questions.

Meeting a friendly electronics boffin in the Rose and Crown the other night, I tackled him on the subject. "Supposing," I said, "I had wired up all the diodes the wrong way round would I get positive ions instead of negative ones?" This particular model works on the Cockcroft Walton voltage doubling circuit. My friendly boffin frowned, and said he would give it some thought and I said I would return later.

When I returned three hours later, he was deep in thought and knee deep in empty glasses. "You see," I told him, "Although my wife finds it makes breathing easier, it might be all in the

mind, in other words, an electronic placebo." I will remember a cousin of mine who had to take a spoonful of medicine each day and got it mixed with my aunt's shampoo. She swore it cured her and we never told her the truth, though my aunt did experience difficulty in trying to wash her hair with the medicament.

The second question is this; my ioniser is working on a low mantelpiece made of ceramic tiles and just under the discharge point, after a few weeks black lines appear which look like a river depicted on a map, thin lines joining thicker ones and finally joining one thicker still. My friend looked alarmed, and hinted that double scotch might help clear his brain. This I provided and finally asked him, would there be a simple way I could test to see if the ions were negative or positive?

His face brightened, "Oh! yes, use an electro-scope". I pointed out that the objection to that was that I didn't have one handy. "Oh! in that case you could comb your hair and produce some static electricity. At this juncture I cut him short, "My dear David, after worrying about writing articles for ten years, I haven't enough hair left to comb!!

That is how the matter stands at the moment, I can only hope that some kind knowledgeable reader will come up with some answers.

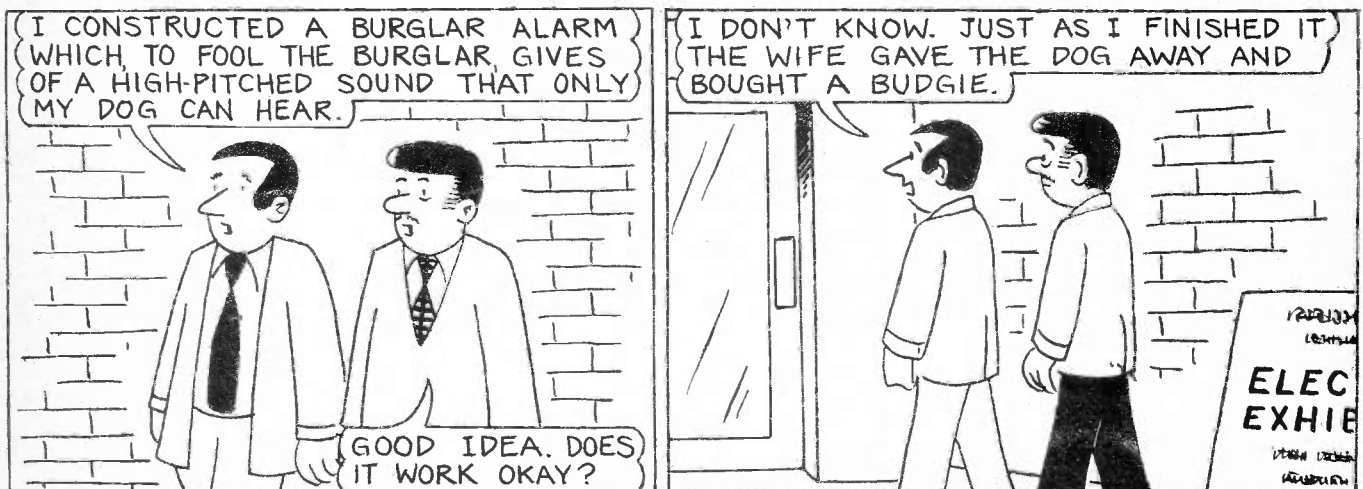
Winds Of Time

One interesting little sidelight on this electronic age came to my attention recently. A teacher who was trying to describe the motion of the wind in a high pressure area said, "It goes in a clockwise direction", Rows of blank faces greeted her remark.

She tried again, "Well look at your watches". The class did this and were still nonplussed. I need hardly tell you that they all wore digital watches.

JACK PLUG & FAMILY...

BY DOUG BAKER



LETTERS

Anti-Phase Phones

Sir—On page 161 of the March 1983 issue of *EE* (*Test Gear 83—Dual Power Supply*), the labels of the resistors R9 and R13 seem to have wandered on to the wrong components.

In Fig. 1c on page 116 of your February 1983 issue (*Square One*), the two halves of a stereo headphone are shown in series to produce a 16 ohm mono instrument by the expedient of connecting to the tip and ring of the jack plug whilst leaving the sleeve unconnected.

Whilst certainly presenting the drive-source with a 16 ohm load, this arrangement will result in the listener's ears being presented with two anti-phase signals which will sound "not right". If the two ear-pieces are to be run in series, the connections to one of them should be interchanged, and this will entail the opening of one of them—not to be recommended.

Personally I have fitted a cheap stereo set with a d.p.d.t. switch so wired that the two sides may either run normally as a stereo set or put in series whilst retaining correct phasing but presenting a 16 ohm mono load, and the result is pleasing.

One final point is that a growing number

of your contributors seem to be specifying scarce and/or special components with RS code-numbers. Now this firm (as you yourselves confirm from time to time) simply will not deign to deal with members of the public (even with chartered electrical engineers like myself), so that the obtaining of specified parts becomes very difficult.

There is the further point that RS are quite merciless so far as continuity of availability is concerned, and are quite likely to drop an item from their list with very little notice. I see their local "rep" every four months or so, and have frequently made this point to him.

For myself, I will not use their products unless they are of simple types available from other sources as well. Perhaps your contributors should be discouraged from specifying parts which are not readily available from genuine retailers.

James W. Robson,
Newcastle Upon Tyne.

Thank you for pointing out the error on the Dual Power Supply project. The values of resistors R9 and R13 were inadvertently transposed on both the circuit diagram and the components' list. It should be R9—24 kilohms and R13—430 ohms. (See Please Take Note, page 304.)

In all fairness to RS Components, it has never been their policy to supply direct to the public, but genuine component distributors and suppliers can usually obtain their product, particularly if a part number is supplied.

As for RS dropping items from their list with little notice, the catalogue does indicate if a particular item is subject to design change or only available until stocks are exhausted and this is always checked before a stock number is quoted.

Any other reader experiencing difficulty obtaining components from suppliers? Comments welcome.

Temperature Sensor

Sir—I read with interest your article "Car Thermometer" by M. Plant in the March issue, and was a little surprised to see such an expensive, precision device as the RS Components 308-809 temperature sensor, at £5.28 trade price, specified in such a design. Then I read the piece in *Shop Talk*.

I feel sure that the National LM334 would be quite as suitable for this thermometer, although it might mean a little redesign. I enclose a brief data sheet on this device, which we do not stock at the moment, but would be prepared to stock for your readers at a price of £2.50 including VAT and postage.

Geoffrey Hillier,
Midland Electronics,
Nottingham.

Interested parties should contact Mr. Hillier at Midland Electronics, Department EE, 5a Gregory Street, Lenton, Nottingham NG7 2LR. We stress that we have not tried this alternative and that some circuit modification is required. A data sheet is available.

A.C. Alternators

Sir—In your series *A.C. Mains Explained* (Part One, page 475, July 1982), Mr. A. Kenyon shows a diagram (Fig. 1.3) of a single phase a.c. alternator with a single conductor rotated in a magnetic field.

I was asked to draw a single phase a.c. generator showing the basic principles and I reproduced, from memory (and it was very close), Mr. Kenyon's drawing and was told that it was completely wrong!

Apparently it should be a loop of rotating conducting material to produce the sinusoidal waveform shown. Could you please explain why Mr. Kenyon shows the conductor as a single conductor passing through the magnetic field when in fact I am told by my instructor that a loop would passably generate "something like two phase a.c." I hope you can help.

Also, why does he show one end to be positive (+) and the other end to be negative (—)?

The series generally has helped me a lot. Thank you.

J. Tustin,
Skillcentre Hostel,
Slough,
Berks.

In order to answer your first enquiry concerning the single conductor rotating in a magnetic field, it is necessary to return to first principles.

In the nineteenth century Michael Faraday discovered the effect of electromagnetic induction, that is, an

e.m.f. (electro-motive force) is induced in a wire when it cuts through a magnetic field.

As Mr. Kenyon stated in his article, the amount of e.m.f. depends on three factors; the strength of the magnetic field, the speed at which the wire moves through the magnetic field and the length of the wire.

If the two ends of the wire are connected so as to make a complete circuit (as in the case of Fig. 1.3, via the slip rings, carbon brushes and the oscilloscope) then the induced e.m.f. will cause a current to flow in the wire. The direction of this current flow (the polarity of the e.m.f.) will depend upon the direction that the wire cuts through the magnetic field. This is known as the Generator Principle.

To return to the example shown in Figs. 1.2 and 1.3, we must think of the magnetic field between the two poles of a permanent magnet as magnetic lines of flux and these are represented on the diagrams as equally spaced parallel lines running from the north pole to the south pole.

If we look at the end-on view shown in Fig. 1.2 with the conductor at point A. At this instant, the conductor is travelling parallel to the flux lines and as such, is not cutting through the magnetic field therefore the e.m.f. is zero.

However, in traversing from point A to B, the conductor starts to cut through the magnetic lines of flux and so generates an e.m.f. At point B it is perpendicular to the field and is therefore cutting through the maxi-

mum number of lines and the e.m.f. reaches its peak value.

Continuing its rotation to point C, the e.m.f. falls back to zero as the conductor is once again travelling in the same direction as the lines of flux. At this moment, the conductor reverses its direction through the magnetic field and so as previously stated, the polarity of the induced e.m.f. and hence the direction of current flow, is reversed.

So as the conductor rotates from point C, through D and back to A, the same e.m.f. is generated as in the first half of its rotation but with the opposite polarity. In this way, a sinusoidal waveform is generated.

So the simple a.c. alternator shown in Fig. 1.3 is in theory correct. However, in practice this arrangement is inadequate to obtain the large amounts of e.m.f. needed for power generation and the practical solution is to rotate a coil of wire (remembering that a "loop" is in fact, just a single-turn coil) in the magnetic field.

In doing this, the length of the conductor is simply being increased, this being the third factor governing the amount of e.m.f. induced.

As to your second question regarding the polarity, by fixing one end of the wire (or loop) at earth potential and calling it OV, the potential at the other end of the wire (with respect to OV) will alternate between a positive peak and a negative peak for each revolution.

Calling these peaks positive and negative simply indicates that the polarity of the e.m.f. has been reversed.

TEXAS DRIVE FOR HOME COMPUTER MARKET

JUST launched in America by Texas Instruments is a new 16-bit Home Computer that is expected to sell for around £75. Unlike most computers in this price bracket, the TI-99/2 Basic Computer can use software on solid-state cartridges as well as on cassettes.

The TI-99/2 is equipped with a staggered QWERTY arrangement, the space bar forming one of the 48 function/operation keys. The computer has 4.2K bytes of random access memory (RAM), of which 4K bytes is user accessible, and can be expanded to a total of 36.2K bytes of RAM.

Most peripherals for the new system will plug into a Hex-bus peripheral interface connector in the rear of the case. The Hex-bus port allows users to connect any peripheral developed for TI's Compact Computer family.

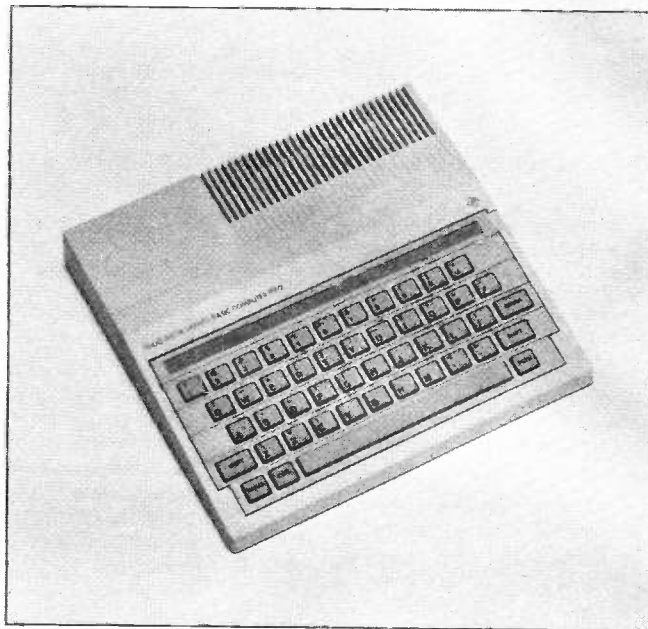
Software

Twenty software programs will be available on cassettes initially but more will be added as they become available. Educational programs include: Picomath-80, Math I and II, Statistics I and II, Sunrise Time, Datetimer and Civil Engineering.

Programs for personal management are: Household formulas, chequebook manager, purchase decisions and general finance. Entertainment cassettes include: Lunar Landing Biplot, The Minotaur, TI Trek, and Mind Games I, II, III and IV.

A built-in r.f. modulator allows connection to any television. A cassette interface cable is also included to interface directly to the new TI program recorder or many of the ordinary tape cassette players on the market.

Availability of the Texas TI-99/2 Basic Computer in Europe is planned for the third quarter of this year. "The TI-99/2 is designed to allow computer novices to learn to program a computer in TI Basic and Basic-supported assembly language," said William Turner, President of the Consumer Group.



COMING EVENTS

A working conference on "Prototyping", sponsored by the Commission of the European Communities will be held in Brussels on 25 to 28 October.

The conference, organised by the National Computing Centre and Gesellschaft für Mathematik and Datenverarbeitung (GMD), Bonn, aims at bridging the gap between researchers and practitioners in the field of software engineering and information systems development.

★ ★ ★ ★ ★

The Essex University "1983 Electronic Systems Summer School for Teachers" will be held from Sunday July 10 to Friday July 15. This year they are offering two revised courses: Feedback and Communication Systems and Digital and Computer Systems.

The Summer School courses have been structured to assist teachers of Electronics within the School environment. In particular, the courses complement the AEB "A" level in Electronic Systems.

However, it is emphasised that both courses have sufficient breadth to be of relevance to similar "A", "O" and CSE curricula.

The first distributor to offer the complete range of Mullard/Signetics "Micromin" components is to be Swift-Sasco. These include ranges of i.c.s, transistors and diodes, all in S.O. packages for surface mounting on printed circuit boards.

The IBA have negotiated a non-exclusive know-how and patent-licensing agreement for the use of its MAC satellite transmission system with the United States Satellite Broadcasting Corporation (USSBC), a firm linked with Hubbard Broadcasting Inc. of Minnesota.

US FAVOURS UK

The UK has consolidated its position as the most favoured overseas location for US electronics investments. As much as 27 per cent of those US electronics companies which plan to set up foreign operations within the next three years have either chosen or shortlisted the UK as their preferred European base—as long as it remains a member of the European Economic Community.

This is one of the findings from the second American Electronics Survey undertaken by "The Electronics Location File". A representative sample of 662 electronics companies were interviewed during November and December 1982 to allow company executives to air their views on issues affecting the US electronics industry and to predict its future performance and domestic and foreign investments for the coming year.

One result to emerge is the almost unanimous belief that the US electronics sector will improve in 1983. More than 80 per cent of the companies surveyed foresaw an improvement and nearly 90 per cent expected their own companies to perform better than in 1982.

Space Communications

State-of-the-Art microwave equipment for use at Goonhilly Downs global satellite communications centre in Cornwall has been supplied by Siemens. The equipment, comprising two high-power amplifiers, automatic carrier level control and a signal routing and switching system, was developed at Siemens engineering laboratory in Congleton, Cheshire.

The two high-power amplifiers are installed in Aerial 2 at Goonhilly and use the Siemens YH 1045 A3 travelling-wave tube, which is also used in high-power amplifiers installed in Hong Kong and Bahrain. Travelling-wave tubes (t.w.t.s) are a wideband alternative to using klystrons for microwave power amplification.

Using t.w.t.s, the carrier signals are combined at low power levels prior to being fed into the t.w.t. for power amplification. Using relatively narrow-band klystrons, on the other hand, carrier signals are amplified individually and then combined at high-power.

Travelling-wave tube high-power amplifiers are, therefore, the last stage in the transmitter immediately prior to the aerial feeder circuits.

FREE SHOPPING BY PRESTEL

Customers of the Nottingham Building Society are being offered free conversion of their television sets to connect with Britain's first electronic banking system. The adaptor and keyboard enables customers of the Nottingham to connect with the Scottish Bank involved in the system, and for the cost of a local telephone call they can check their account, transfer monies as well as being able to perform other financial actions.

The conversion enables customers to use the Prestel computerised system and it is hoped retailers will soon be connected to extend the service with "Home Shopping". The consortium hopes to attract 100,000 customers by 1986 and sees this scheme as the first major competition for the clearing banks when the "Home Shopping" service enables customers to order, and pay for goods, debiting their accounts with the Nottingham by computer.

Prestel are financing half the cost of each installation which is expected to total £100 per household.

Despite a loss of \$1.6M in absorbing the Stromberg-Carlson acquisition, Plessey has reported an 18 per cent increase in third-quarter profits to £29M.

This means that for its first nine months Plessey has increased its profits by 14 per cent to \$102M and is well on target to surpass the 1982 end-of-year figure of \$111.4M.

Power Approval

The first UK power supply manufacturer to gain the critical UL, CSA and VDE approvals on a standard catalogue range of switch mode power supplies has been awarded to Gould's of Bishop Stortford.

The approvals, which are accepted internationally as guaranteeing the electrical safety of sub assemblies designed to be included in business machines or consumer equipment, apply to all units in Gould's Simflex open-frame range, which is aimed at the microcomputer and business system market.

Transatlantic Lightlines

Tenders are now being invited for the world's first inter-continental optical fibre under-sea cable. It will be able to carry many thousands of phone calls across the Atlantic as pulses of laser light in strands of glass no thicker than a human hair.

The new cable, due to start service in 1988, is expected to cost more than £250 million. British Telecom will contribute the second largest share of this sum.

Cable TV Has Little to Offer

Delegates to a recent NCC Workshop on the business applications of Cable TV Systems concluded that cable systems had little to offer the business users.

With the high speed 2Mbit/s and 8Mbit/s digital circuits becoming available from British Telecom and Mercury, little need was seen for the still higher capacity of cable systems.

Business users generally require a large number of incoming and outgoing circuits and this need is not compatible with cable systems, where incoming circuits predominate. Nevertheless, delegates did feel that to cater for possible future business needs, there should be a unified national addressing scheme for terminals on cable systems to give every terminal a unique address.

BREAKFAST IN TIME

To accommodate the launching of "Breakfast Time TV", the BBC Topical Production Centre at Lime Grove underwent a major facelift in record time. The major change being the construction of a new three-storey technical area and production offices within a former film studio.

Production Area

The "Breakfast Time" production offices incorporate a Hewlett Packard 3000 computer, which is used to assist the production staff in the preparation of stories, scripts and running orders. The majority of the hardware is located in an old dressing room, with groups of terminals connected to it via fibre optic cables; fibre optics were used to avoid electrical interference to data signals, and to simplify cable runs.

The 40 terminals can be used as word processors to prepare new material or may have news agency wire services outputs displayed direct on the screen. When stories have been prepared they can be sent out automatically to the programme editor; details such as the presenter's name, the running time, and notes of any videotape or telecine inserts can be added. The final script is then available to the studio director on a VDU in the studio control room, or can be printed out for use by the presenters. The computer project has been made possible by a grant of £250,000 from the Department of Industry during the 1982 Information Technology year.

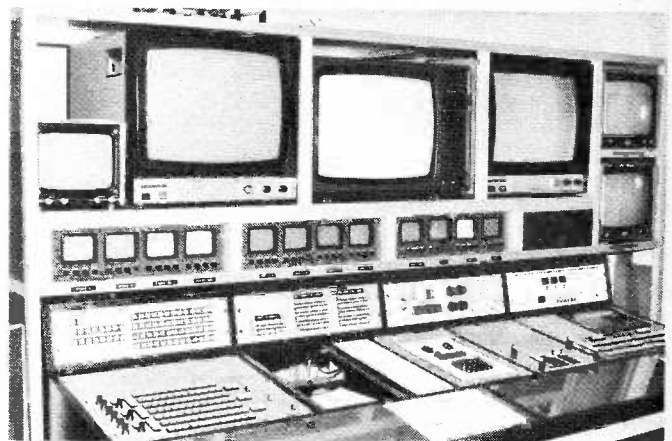
Technical Base

The new technical area combines videotape (VT), telecine

(TK) and electronics graphics equipment in a coherent group, and is adjacent to the Central Vision Apparatus Room (CVAR). Equipment recently added in the CVAR has included video, audio and d.c. matrices. The five matrices allow outside sources to be connected to the VT area, VT and TK to be fed to the studio mixers and out-going lines, monitoring of Network and Test Signals, cue and talkback routing,

and connections to and from the electronics graphics area.

The electronics graphics area is next to the new VT area. It has been equipped with a Quantel 6001 Stills-Store (which has four 180M byte disc packs to enable over 700 stills to be stored), a Quantel 7001 paint box computer graphics system as well as a two-channel Chyron IV caption generator, and a simple video rostrum equipped with an Ikegami HL79D camera. A control desk contains an effects mixer to combine the outputs of the various graphics devices, the matrix controls to route the graphics devices to either studio or to VTR machines, and comprehensive monitoring.



New facilities at the BBC's Topical Production Centre at Lime Grove include an electronic graphic artists area. Control of the Quantel paint box and Stills-Store is via this effects mixer and matrix.

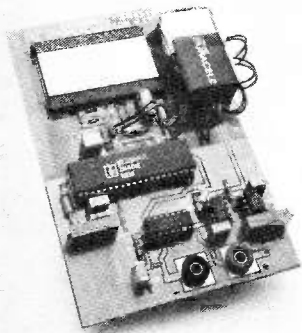
NEW · NEW · NEW · NEW

PRODUCTS

NEW · NEW · NEW · NEW

DVM EVALUATION KIT

AN EVALUATION kit for the recently released **Ferranti ZN451** Digital Voltmeter is now available. It is claimed that the customer can evaluate the capabilities of this 3½ digit monolithic DVM without the need to resort to designing and constructing a system from scratch.



One of the features of the ZN451 is the facility whereby external circuits may be included in the auto zero loop: output signals are provided to control external auto zero switches so that op. amps or other signal conditioning circuits can be included in the loop to boost input impedances and/or improve sensitivity to as low as 1.999mV full scale. Thus it is possible to measure low voltages without any "zero error".

Full details of both the Evaluation Kit, which is priced at £29.50 and the ZN451 are available from:

Ferranti Electronics Ltd.,
Dept EE, Fields New Road,
Chadderton, Oldham,
Lancs OL9 8NP.

TALKING COMPUTERS

A NEW speech synthesiser device that allows computers to speak whatever is typed-in on a keyboard, is now available from **Intelligent Artefacts**, a division of Sands Whiteley of Orwell, Royston, Herts. Known as the **Votrax Type-n-Talk**, it can be used with practically any available computer. It is claimed to have unlimited vocabulary and allows computers to talk with a reasonably clear "voice".

It is operated by simply typing English text and a talk command. Typewritten words are automatically translated into electronic speech or code by the unit's microprocessor-based text-to-speech algorithm.

It is possible to program verbal reminders to prompt you after a lengthy routine and make your computer announce events. Computer games can be made to speak of dangers, give reminders, and praise the contestant.

In teaching, the computer with **Type-n-Talk** can be used in spelling tests. For the blind it allows them to undertake computer studies and it can help the dumb with speech output. For example: by typing in the characters representing "h-e-l-l-o", the spoken word "hello" is generated.

The synthesiser was developed and produced by **Votrax** of Detroit, USA, for whom **Sands Whiteley** are exclusive European distributors. It is available for £275 plus VAT.



SOLDERING STYLUS

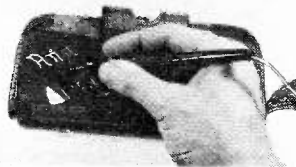
A NOVEL and useful marking tool has just been marketed by **Light Soldering Developments**. The **Admin Electric Stylus** is about the size of a ballpoint pen, and operates at 4½ volts from its own mains plug/isolation transformer, which fits a standard 13A socket.

The silver-alloy writing tip heats in about 30 seconds to a temperature (sufficient to activate the special blocking foil supplied), to allow writing in a choice of metallic and coloured finishes on surfaces such as paper, card, plastics, leather and paint. This makes it ideal for modelling and for lettering front fascia panels on equipment.

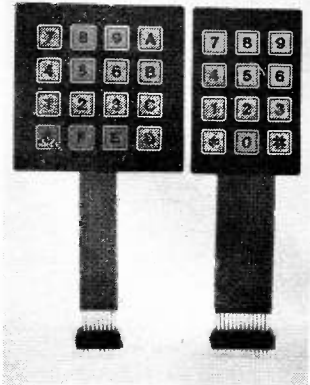
The **Admin Stylus** comes complete with plug/transformer, 9 lengths of foil (gold, silver, copper and six colours) and full operating instructions. The recommended retail price is £15.04, including postage and VAT.

For more details and addresses of local stockists contact:

Light Soldering Developments Ltd.,
Dept EE, 97/99 Gloucester
Road, Croydon
CRO 2DN.



KEYBOARDS



A NEW range of membrane keypads, available with 12 or 16 keys, has just been announced by **Velleman (UK) Ltd.**

Both versions, type **KB12** and **KB16**, are offered with standard legend or blank keys, to enable the user to make up their own legend. Insulated flat ribbon cable is used to connect the keyboard to a 0.1in. spacing p.c.b. connector. Ratings are 24V and 25mA maximum.

Price including VAT and postage is £8.44 for both versions, on one off quantities, with discounts available on large orders. They will also manufacture keyboards to customers' own design.

A data sheet with full technical specification is available on request.

Velleman (UK) Ltd.,
Dept EE, P.O. Box 30,
St. Leonards-on-Sea,
Sussex TN37 7NL.

TRANSFORMERS FOR DESIGNERS

A NEW customer design facility aimed at the circuit designer has been set up by **ILP Electronics**, the audio power amplifier module people, for the supply of "special one-off" toroidal transformers.

They offer a seven day design and prototype service, for which they charge a nominal £20 on top of the one off price. All custom design transformers are allocated a part number upon manufacture of the prototype.

ILP have also appointed **Barrie Electronics Ltd.**, of 3 The Minories, London EC3, as one of their distributors and stockists of the new toroidal transformers.

For details of the new service readers should write to:

ILP Electronics Ltd.,
Dept EE, Graham Bell House,
Roper Close, Canterbury,
Kent CT2 7EP.



SOLAR SCIENTIFIC CALCULATOR

EXPERTISE in calculator miniaturisation and solar cells have been brought together in the new **Casio FX98**, a no batteries credit card sized calculator with 42 scientific functions.

A bank of solar cells located in the top corner power unit from sunrise to sunset. The liquid crystal display gives an eight-digit mantissa and two-digit exponent readout. A full set of log, trig, power and root functions are possible, plus statistical analysis, ability to handle fractions, polar-rectangular and sexagesimal conversion.

The Casio FX98 solar scientific calculator has a recommended retail price of £16.95 and addresses of local stockists can be obtained from:

**Casio Electronics Ltd.,
Dept EE, Unit 6,
1000 North Circular Road,
London NW2 7JD.**



A LOGICAL KIT

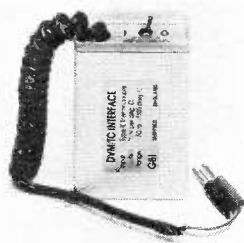
AIMED originally at school pupils studying for 'O' and 'A' levels, the Digital And Logical Electronics Circuit Kit from Carter-Richardson is equally suitable for anyone looking for a "starter" course in logic and digital electronics.

Claimed to be designed by practising teachers of electronics, the Daleck kit consists of eight interlinking modules together with a progressive guide to experiments. These circuit modules are also available separately.

Standard TTL 74 series integrated circuits are used to teach the properties of the standard gates and combinational circuits. The properties of the different flip-flops are demonstrated and sequential circuits for serial and parallel registers and accumulators are developed. Indications are also given on how to extend the system into other areas.

Carter-Richardson Electronic Systems, Dept EE, Greta Side, Keswick, Cumbria CA12 5LG.

TEMPERATURE INTERFACE FOR DVM



ANYONE who has access to a digital multimeter can now use it as a wide range temperature measuring instrument using standard type K thermocouples, by adding the DVM/TC Interface Unit from **Graham Bell Instrumentation**.

This new product has a temperature range of -50°C to $+1100^{\circ}\text{C}$ and incorporates automatic cold junction compensation. The output of 1mV per degree Centigrade is via a coiled lead fitted with 4mm plugs.

Since the accuracy is claimed to be unaffected by the output loading, it may also be used to interface low output impedance instruments such as chart recorders.

Further information can be obtained from:

**Graham Bell Instrumentation,
Dept EE, P.O. Box 230,
39 Derbyshire Lane,
Sheffield S8 0TH.**

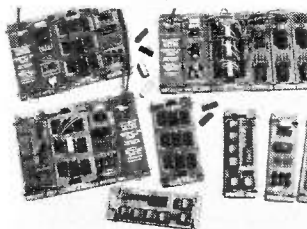
ON FILE

A SELECTION of needle files has just been announced by Neill Tools and should prove invaluable additions to the tool box.

A set of six 16mm files in a plastics wallet consist of one each: hand, flat, round, half-round, square and three square.

These files are ideal for clearing copper burrs across pads on p.c.b.s and solder "whiskers" between stripboard tracks. The recommended retail selling price is £8.22, excluding VAT.

**Neill Tools Ltd., Dept EE,
Napier Street, Sheffield
S11 8HB.**

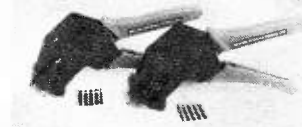


RIBBON CABLE STRIPPER

OVER the last few months we have published several projects that have called for the use of ribbon cable for interwiring boards, controls and interface connectors. The stripping or baring the ends of this type of cable can be fairly tricky and any tool which aids this delicate operation is most welcome.

Just such a tool has come to our notice with the announcement from **Eraser International** of two new ribbon cable stripping pliers in their range of production equipment.

The models RTS1 and RTS2 both have an adjustable strip length incorporated and are adjustable for different cable sizes



and self-adjusting for cable widths.

The model RTS1 will remove the insulation completely from the end of a cable but the RTS2 is designed to give only a partial strip, thus leaving the strands intact until they are required for termination purposes.

The recommended retail price is £95 and further information is available from:

**Eraser International Ltd.,
Dept EE, Unit M, Portway
Industrial Estate, Andover,
Hants SP10 3LU.**

EVERYDAY ELECTRONICS COMPUTER PROJECTS

Published over the last 12 months

- 2K RAM PACK FOR SINCLAIR ZX81 *April 82*
- KEYBOARD SOUNDER FOR SINCLAIR ZX81 *June 82*
- TEMPERATURE INTERFACE FOR TANDY TRS-80
Aug/Sept 82
- EXPANSION SYSTEM FOR SINCLAIR ZX81 *October 82*
- TAPE CONTROLLER FOR SINCLAIR ZX81 & SPECTRUM
November 82
- EXTRA RAM FOR SINCLAIR ZX81 *December 82*
- A TO D CONVERTER FOR PET *January 83*
- SPEED COMPUTING SYSTEM FOR SINCLAIR ZX81
February 83
- EPROM PROGRAMMER FOR ACORN ATOM *Feb/March 83*
- EXPANDED ADD-ON KEYBOARD FOR SINCLAIR ZX81
March/April 83
- AMPLIFIER FOR SINCLAIR SPECTRUM *April 83*

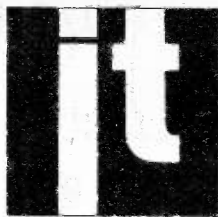
All the above are fully illustrated, detailed constructional articles. Back numbers £1.00 (inclusive of p & p) per copy currently available from: Post Sales Department, IPC Magazines Ltd., Lavington House, 25 Lavington Street, London SE1 0PF.

Please enclose appropriate remittance.

Stocks of back issues are limited. In the event of non-availability a photocopy of the relevant article will be supplied.

Software Service

This new EE Software Service starts this month. It provides the user with a quality copy of the master operating program for our computer based projects. Programs are available on cassette tape. See page 281 for further details.



PART SEVEN

BY T.E. IVALL C.Eng., M.I.E.R.E

In the first article of this series a fundamental question was posed about information technology: is **IT** a genuinely separate engineering activity which we shall continue to recognise as such long after 1982's *Information Technology Year* is over? And after looking at some of the activities and products involved we asked a subsidiary question: is there really something essential which holds all these disparate elements together in some kind of natural unity?

The time has come to round off the series by summing-up and offering answers to these questions—or perhaps by admitting that no satisfactory answers can be found.

PUBLIC AWARENESS

Well, if public events and the mere repetition of the name are anything to go by, it certainly seems that **IT** exists. At the end of the government's **IT82** public awareness campaign the Prime Minister announced at a London conference that "according to the results of an opinion poll I've just received, six out of ten people have heard of **IT**. At the beginning of the year fewer than two in ten had heard of **IT**."

Of course, this proves nothing beyond the effectiveness of a publicity campaign. The reaction of many intelligent citizens seems to be: "Yes, I've now heard of information technology but I still don't know what it really is". At one point the advertising campaign was implying that **IT** was just a new name for computers. Those hoping for enlightenment must have been confused.

More down to earth were the various comments coming from the industrial world. In October 1982, for example, a government appointed committee chaired by Mr J. A. Alvey of British Telecom reported on "A Programme for Advanced Information Technology". In response to Japanese competition they had been asked to advise on a British programme of collaborative research and development into **IT**. The report said there was a need for such a national effort and identified four areas which were felt to be very important: software engineering; v.l.s.i. (very large scale integration); man-machine interfaces; and "intelligent knowledge-based systems" (using inference in the performance of industrial tasks).

Then in February of 1983 the National Economic Development Office made parallel recommendations, for production

rather than R&D, in a report entitled: "A Policy for the United Kingdom Information Technology Industry". As far as the UK is concerned NEDO says that **IT** is alive but not at all well. We have a trade deficit of £150M a year which may grow to £1000M a year by 1990 if we don't do something about it.

AN AREA FOR STUDY

But more pertinent to our questions, perhaps, is the news that academia is recognising **IT** as an identifiable area for study. For example, the University of Surrey has just established a Chair of Information Technology, sponsored jointly by Racal Electronics and the Department of Industry. And the universities of Aston and York are developing new degree courses, with the help of British Telecom, for training electronics engineers to become skilled in both telecommunications and computers—the two main arms of **IT** (see *Everyday News*, March issue).

So it looks as though the academic world has found **IT** real enough to be worth teaching as a distinct, if not entirely separate, subject. This follows a pattern of increasing specialisation which has already given us new university departments for such things as Cybernetics and Computer Science.

THE BINARY CHOICE

All this is really too recent to allow us to digest it properly and perhaps find some answers in it. Meanwhile, as far as the present author is concerned, it still seems that any unity to be found in information technology must depend on the fundamental principle of the binary choice. In technological terms this comes down to digital electronics.

The binary choice is common to the binary logic of digital computers (Part 6) and the binary coded information transmitted in digital telecommunication systems (Parts 2 and 4). Electronic logic is concerned with the information of states; digital transmission systems with the information of coded numbers and characters.

If history is anything to judge by, these two methods go back a very long way. Deductive logic with its "true" and "false" statements, as we saw in Part 6, has been practised for many centuries, perhaps originating with Aristotle. Binary coding of characters, as George Hylton showed in the March issue (p. 146), goes back at least as far as Francis Bacon, the

Elizabethan philosopher and man of affairs. It's probably a very ancient principle indeed if we also think of smoke signals or the high and low pitch drum-beats of the bush telegraph.

DEEPLY ROOTED

So history shows the basic concept of the binary choice to be deeply rooted in human thinking. Thoughts often lead to deeds. What we are now doing in information technology is just a mechanisation of this fundamental idea for useful purposes. The significant fact, mentioned in Part 1, that the binary digits of transmitted codes are sometimes generated as the binary states of electronic logic, linking telecommunications with computers, is one technological example of how central is this principle of two-state representation.

● But what do *you* think? **EVERYDAY ELECTRONICS** would be glad to hear your views on **IT** and the Editor would certainly be happy to publish any letters with interesting contributions to this discussion. □

PRACTICAL ELECTRONICS

Our Sister Publication

PRACTICAL ELECTRONICS features the following projects in the May issue.

PROJECTS

- ZEAKER**—Inexpensive buggy type robot for your ZX81
- AUTOMOBILE TEST SET**
- PERSONAL STEREO AMPLIFIER**
- MAINS WATCH DOG**

FEATURES

- SEMICONDUCTOR CIRCUITS**—using chips—
- NEW REGULAR FEATURE INTO THE REAL WORLD**—
- Intertacing micros

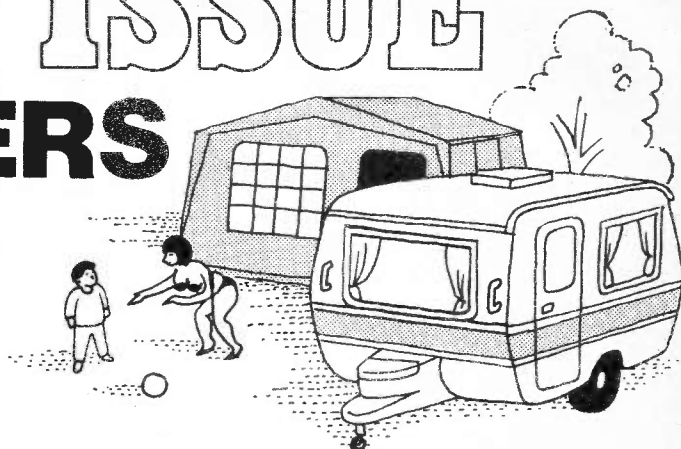
*Don't miss this bigger issue—
more pages—more coverage—
more value*

ON SALE NOW

JUNE ISSUE

CARAVANNERS & CAMPERS

HERE ARE TWO SIMPLE YET EXTREMELY VALUABLE UNITS THAT WILL HELP ENSURE MAINTENANCE OF AMENITIES CONSIDERED NORMAL NOWADAYS AND EXPECTED WHEN ON TOUR OR LOCATED AT A HOLIDAY SITE.



CARAVAN POWER SUPPLY

A simple and effective unit to enable an auxiliary battery in the caravan to be re-charged on a touring holiday thus enabling interior accessories to be used independently of the towing vehicle. Not designed for long-stay holidays at fixed locations but invaluable to the tourist.

CARAVAN FRIDGE ALARM

For use with absorption fridges of the type used in caravans and for camping. Gives audible or visible early warning of fridge failure. Fully effective when towing or at a fixed site. Independently powered from PP3 battery.

TRS-80 & VIDEO GENIE MICRO USERS

EPROM PROGRAMMER

A PORT-BASED SYSTEM (THREE 8-BIT OUT PORTS, ONE 8-BIT IN PORT) DESIGNED FOR USE WITH A HOST COMPUTER TO PROGRAM 1K, 2K AND 4K +5V RAIL EPROMS. SOFTWARE DEVELOPED FOR USE WITH THE TRS-80 MODEL I LEVEL II AND VIDEO GENIE MICROCOMPUTERS, BUT MAY BE USED WITH OTHER MACHINES PROVIDING THAT SUITABLE SOFTWARE IS DEVELOPED AND CERTAIN SYSTEM SIGNALS ARE ACCESSIBLE.

GUITAR EFFECTS UNIT

PUSH BUTTON COMBINATION LOCK

PLUS ALL THE REGULAR FEATURES

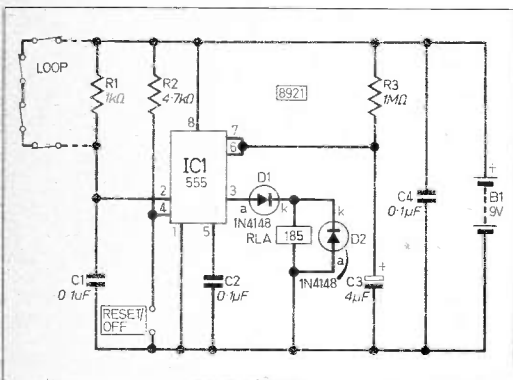
EVERYDAY
ELECTRONICS
and computer PROJECTS

JUNE 1983 ISSUE ON SALE FRIDAY, MAY 20

CIRCUIT EXCHANGE

BURGLAR ALARM

THE heart of the circuit lies around the 555 timer i.c. Once the alarm has been triggered it will latch on even if the loop is remade. If the alarm has been triggered and the loop remade, it will turn off automatically after a preset period depending on the value of R3. The value specified was 1 megohm which gives a delay of about one minute and six seconds.



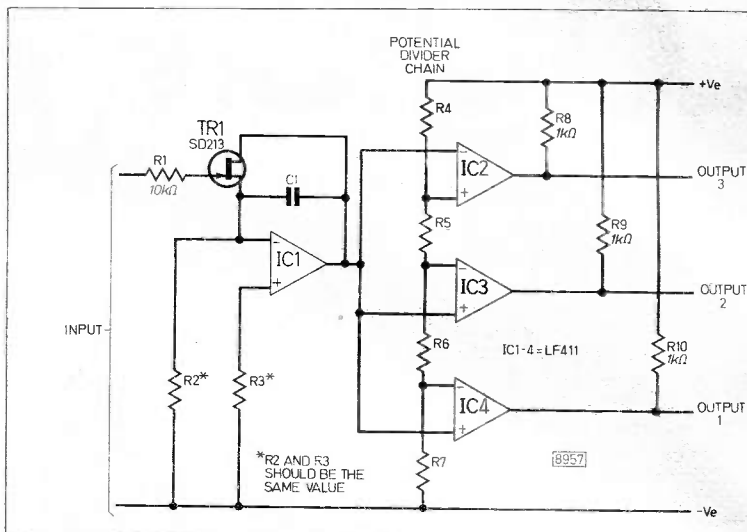
The 555 is triggered by pin 2 of the i.c. being held positive when the loop is made. When it is broken pin 2 becomes negative thus triggering off the 555.

It was found that C1 and C4 were necessary as these suppressed any interference picked up by the loop which might accidentally trigger the alarm. The circuit will run from 5V to 15V d.c. depending on which kind of relay is used.

S. Currell,
Hoveton,
Norwich.

VARIABLE SEQUENTIAL SWITCH

A POSITIVE voltage is applied to the input and therefore also the gate of TR1, the f.e.t. TR1 is turned on and C1 appears to be short circuit. When the input voltage is reduced to 0 the short is removed and IC1 acts as an integrator.

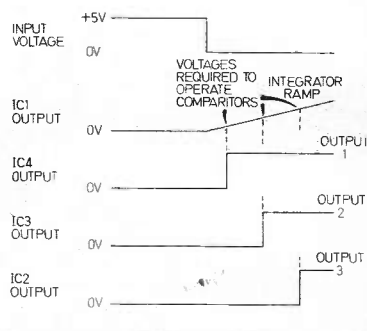


An enhancement mode insulated gate f.e.t. (TR1) is used to ensure that when the input is 0 there is no leakage, that is no parallel path across C1.

The integrator ramp rate = $(-ve/R2 C1)$ volts per second. Operational amplifiers IC2, IC3, IC4 are comparators. When the ramp voltage on the non-inverting input exceeds the voltage on the inverting input the comparators change state and an output is generated. The voltage at the inverting input is derived from the potential divider chain. As each position on the divider chain will have a different voltage available, so each

comparator will operate in sequence but for different levels of ramp input voltage. Because of the resistors connected to the output of each comparator, once the comparators have toggled they will generate the same output voltage.

D. A. Fownes,
Pendeford,
Wolverhampton.



PLEASE TAKE NOTE

Test Gear 83—Dual Power Supply (March 1983)

The circuit diagram (Fig. 1, page 160) and the Components' List (page 161) show the values of resistors R9 and R13 transposed.

They should read R9 as 24kΩ and R13 as 430Ω. The positions of these two components on the circuit board layout (Fig. 3, page 162) are correct.

Test Gear 83—Function Generator (April 1983)

Note: R11 on the circuit diagram should be 470Ω. Fig. 2. Lead marked SK4 should read VR4.

BICC Vero

SPECIAL OFFER COUPON

In conjunction with Everyday Electronics Real Time clock computer module project - BICC-Vero Electronics are offering **£2 OFF** the 'Apple' m/board (order code 200-22266B OR **£1 OFF** the 'Microboard' (order code 200-22271B)

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LP-1 Logic Probe

The LP-1 has a minimum detachable pulse width of 50 nanoseconds and maximum input frequency of 10MHz. This 100 K ohm probe is an inexpensive workhorse for any shop, lab or field service tool kit. It detects high-speed pulse trains or one-shot events and stores pulse or level transitions, replacing separate level detectors, pulse detectors, pulse stretchers and pulse memory devices.

All for less than the price of a DVM

£31.00*

Model LP-3 illustrated

LP-2 Logic Probe

The LP-2 performs the same basic functions as the LP-1, but, for slower-speed circuits and without pulse memory capability. Handling a minimum pulse width of 300 nanoseconds, this 300 K ohm probe is the economical way to test circuits up to 1.5 MHz. It detects pulse trains or single-shot events in TTL, DTL, HTL and CMOS circuits,

replacing separate pulse detectors, pulse stretchers and mode state analysers.

(Available in kit form LPK-1 £13.25)

£18.00*

Model LP-3 illustrated

*Price excluding P&P and 15% VAT

LP-3 Logic Probe

Our LP-3 has all the features of the LP-1 plus extra high speed. It captures pulses as narrow as 10 nanoseconds, and monitors pulse trains to over 50 MHz. Giving you the essential capabilities of a high-quality memory scope

at 1/1000th the cost. LP-3 captures one shot or low-rep-events all-but-impossible to detect any other way.

All without the weight, bulk, inconvenience and power consumption of conventional methods.

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Model LP-3 illustrated

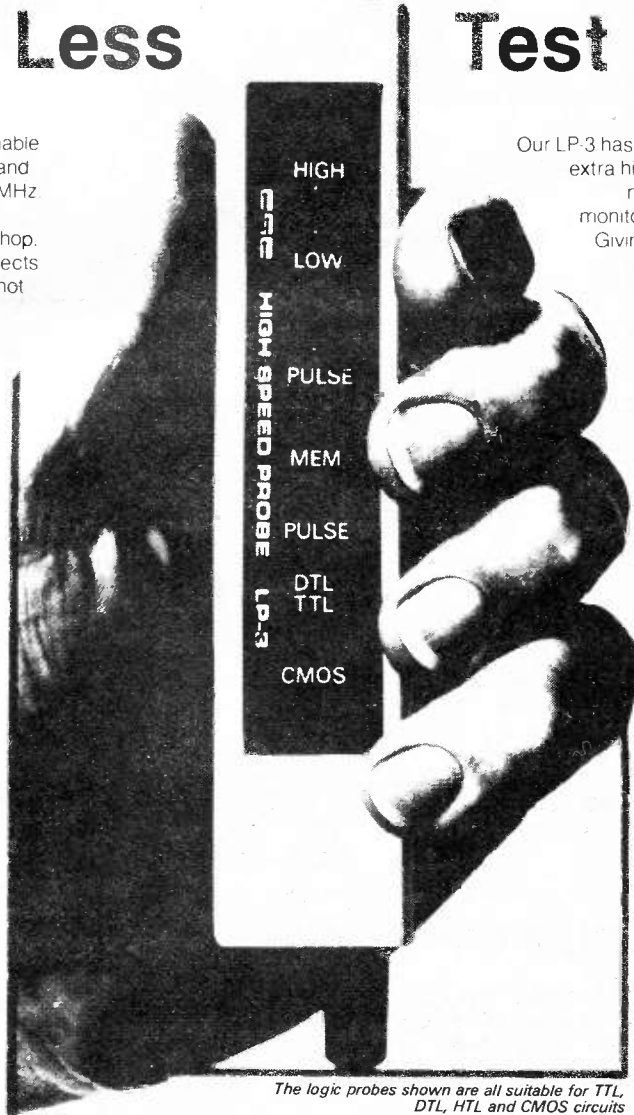
The New Pulser DP-1

The Digital Pulser, another new idea from G.S.C. The DP-1 registers the polarity of any pin, pad or component and then, when you touch the 'PULSE' button delivers a single no-bounce pulse to swing the logic state the other way. Or if you hold the button down for more than a second, the DP-1 shoots out pulse after pulse at 1000 Hz.

The single LED blinks for each single pulse, or glows during a pulse train. If your circuit is a very fast one, you can open the clock line and take it through its function step by step, at single pulse rate or at 100 per second. Clever! And at a very reasonable price.

£51.00*

Model LP-3 illustrated



The logic probes shown are all suitable for TTL, DTL, HTL and CMOS circuits

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G.S.C. (UK) Limited, Dept. 411
Unit 1, Shire Hill Industrial Estate,
Saffron Walden, Essex CB11 3AQ.
Telephone: Saffron Walden (0799) 21682.
Telex: 817477.

G.S.C. (UK) Limited, Dept 411, Unit 1, Shire Industrial Estate, Saffron Walden, Essex CB11 3AQ.
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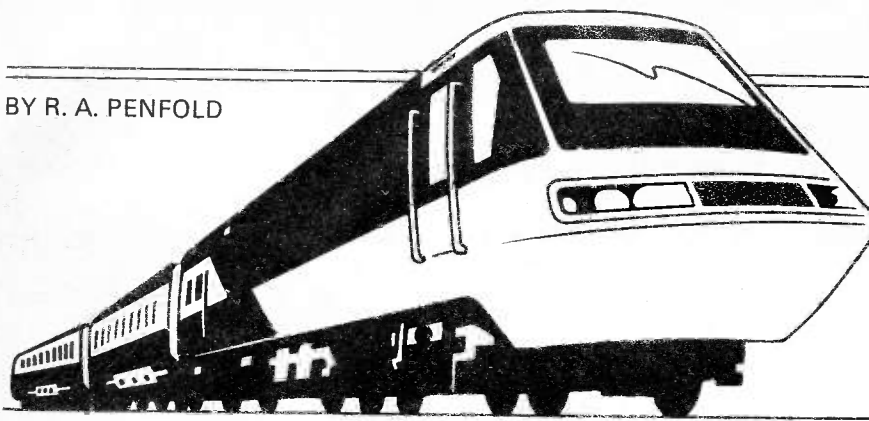
LP-1	£37.38	Qty	LP-2	£22.42	Qty	LP-3	£58.08	Qty	DP-1	£60.95	Qty	LPK-1	£16.39	Qty
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MODEL TRAIN CONTROLLER

VERY simple train controllers are attractive if a low cost controller is required, but they suffer from some drawbacks. One of these is simply that they are rather unrealistic in that the train tends to respond instantly to changes in the speed control setting.

Real trains start much more sluggishly due to inertia, and once underway tend to coast for some distance if the throttle is cut back to zero, due to the momentum of the train. Brakes are needed in order to bring the train to a halt fairly rapidly.

These effects can be simulated electronically by having a speed control that has delays which permit only a relatively gradual increase in output power, and an

even slower decay in the output power. A second control which enables the decay time to be shortened gives the simulated braking action.

SPEED REGULATION

Another drawback of simple controllers is that they give only poor starting performance and speed regulation. This is due to the fairly high output impedance of these controllers, especially at low output power settings.

When an electric motor is stationary it has a very low impedance, and it virtually short circuits the output of the controller. Coupled with the rather high output im-

pedance of the controller this gives very little voltage across the motor, and hence only a low power in the motor.

When an electric motor starts to operate, its impedance increases, and if it is fed from a rather high impedance source this results in the voltage and power fed to the motor suddenly rising by a considerable amount.

In practice this gives an undesirable effect in the form of the train almost instantly reaching a fairly high speed as it starts off, giving little realism. However carefully the speed control is adjusted, this sudden start is still obtained.

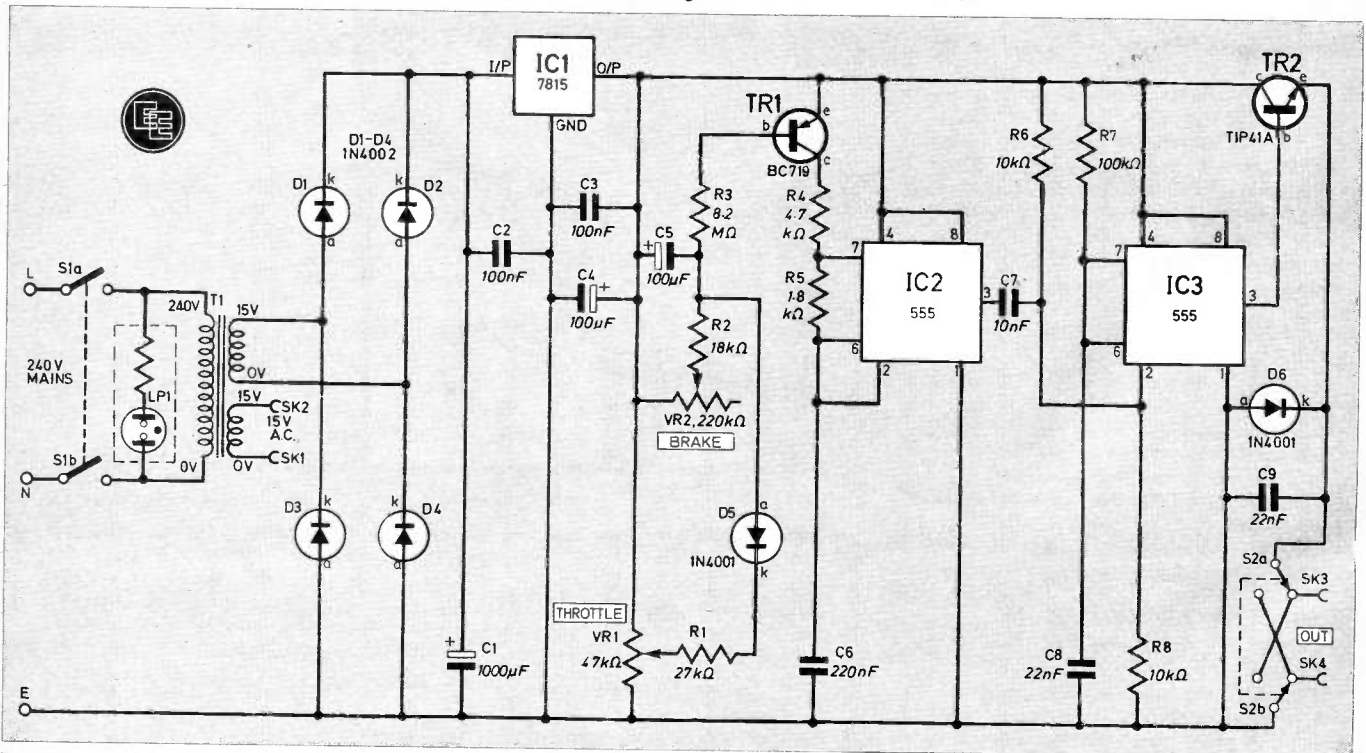
The poor speed regulation is caused in a similar way. The impedance of the motor tends to rise as the train runs down a gradient and the loading of the motor decreases, while the impedance reduces as the train goes up a gradient and the loading on the motor increases. This tends to give more voltage and power across the motor as the train goes downhill, and less voltage and power as the train goes uphill.

This is obviously the opposite of what is needed in order to give good speed stability, and when operating at low speeds the train accelerates down gradients and slows right down or stalls when climbing gradients.

IMPROVED PERFORMANCE

There are several ways of obtaining improved performance, and the most simple of these is to use a controller that has a low output impedance. This gives a stable voltage across the motor so that it draws more current and power when loading is increased, and less current and power when loading is decreased. This

Fig. 1. Complete circuit diagram of the Train Controller.



helps to combat speed fluctuations and stalling, but still gives less than perfect performance.

Even better performance can be obtained by using a circuit that actually gives slightly increased voltage when loading increases, and decreased voltage when loading reduces. In order to give really good results a circuit of this type needs to be matched to the motor it is powering, and this system is not often used in train controllers.

A system that is used a great deal in train controllers, and gives really excellent results is the pulsed controller system. The output voltage is either zero or maximum with this system, except for the very brief periods when the output signal is making the transition from one state to the other.

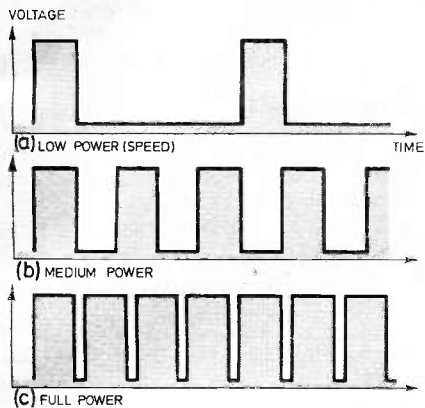


Fig. 2. Output waveform diagrams.

The effective output voltage is varied by altering the proportion of the time for which the output is at maximum. For example, if the controller has an output waveform of the type shown in Fig. 2(a) the output is at zero for a much longer time than it is at maximum. This gives a low average output voltage and thus low power from the motor. In the waveform of Fig. 2(b) the maximum and minimum output times are identical, and this gives half power from the motor. The waveform of Fig. 2(c) gives an output potential that is at maximum for by far the majority of the time, and therefore gives virtually full power from the motor.

PULSED CONTROLLER SYSTEM

The point of this system is that it gives short bursts of high power from the motor when set for low output powers. These bursts of maximum power give high torque from the motor while they are present, and this helps to give the "kicks" that are needed in order to get the train started. They also resist any tendency for the train to stall or run away down hills.

Ideally then, a train controller should provide simulated inertia, momentum, and braking, and should be a pulsed controller circuit. Few circuits of this type have been produced, and most are rather complex and expensive. The unit described here provides all the features

mentioned above, gives very realistic results, but is nevertheless reasonably simple and inexpensive to construct.

There are two ways of generating the control pulses; one way is to have a monostable multivibrator which gives an output pulse of fixed duration each time it is triggered, and then trigger it from a variable frequency oscillator.

A low oscillator frequency gives few output pulses and low output power, a high oscillator frequency gives many output pulses and high power. The alternative method is to have a fixed oscillator frequency and vary the pulse length to control the output power.

VOLTAGE CONTROLLED

This unit uses the former method. The oscillator is a voltage controlled type, and the speed of the train is therefore altered by varying the control voltage fed to the v.c.o. (voltage controlled oscillator).

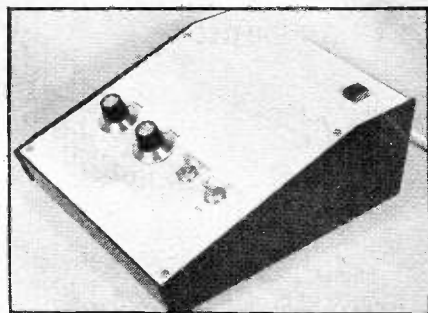
The control voltage is derived using a potentiometer, but the output of this is fed to the v.c.o. via a delay circuit. This gives the necessary short delay when the speed control is advanced, and longer delay time when it is backed-off. The brake control is also incorporated into this part of the circuit.

A buffer stage is needed at the output of the unit in order to provide the fairly high operating current of a model locomotive motor.

CIRCUIT DESCRIPTION

The relatively low cost of the unit is made possible by building the circuit round two 555 timer i.c.s. Fig. 1 shows the full circuit diagram of the unit.

The circuitry around T1 and IC1 forms a conventional stabilised supply capable of delivering up to 1A at 15 volts. IC1 incorporates current limiting cir-



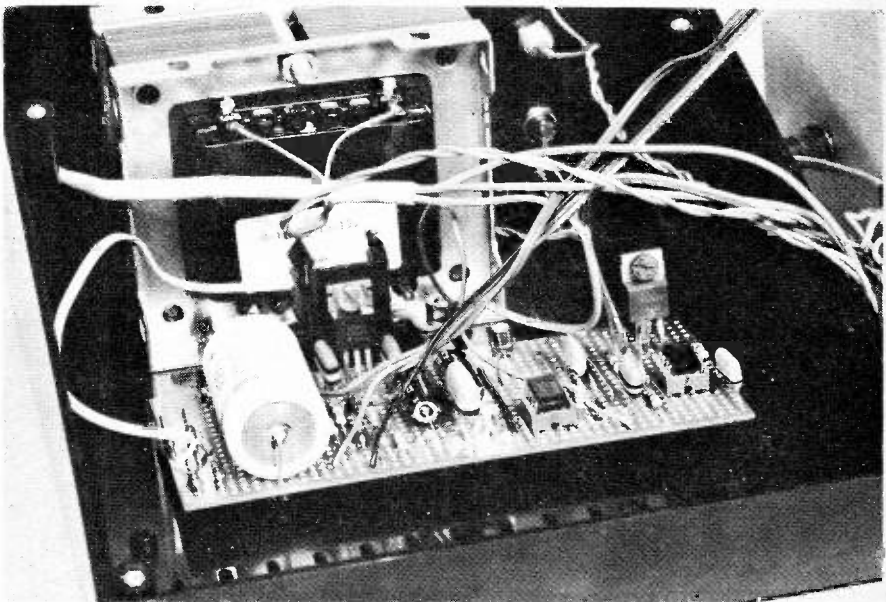
Front view of case showing controls.

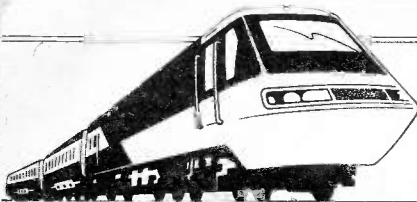
cuitry which protects the unit against damage when the inevitable short circuits on the output occur. T1 has two secondary windings, one of which is used to power the controller circuitry, and the other is fed direct to a pair of output sockets. This gives a 15 volts 2A a.c. output which can be used to power points, controllers and other accessories.

The output pulses are generated by IC3 which is a straightforward 555 monostable circuit. TR2 is used as an emitter follower output stage which gives the unit a suitably low output impedance. R6 and R8 bias the trigger input of IC3 to about half the supply voltage, and the monostable can be triggered by reducing this voltage to less than one third of the supply potential. The trigger pulses must be very brief or they will significantly elongate the output pulses and adversely affect the performance of the circuit.

IC2 is a 555 device used in the astable (oscillator) mode. C7 couples the output of IC2 to the trigger input of IC3 and gives the required brief trigger pulses on the negative transitions of IC2's output. The operating frequency of IC2 is largely controlled by the emitter-collector resistance of TR1. This resistance can be

Complete wiring details of component board mounted inside the case.





MODEL TRAIN CONTROLLER

COMPONENTS

Resistors

R1	27kΩ
R2	18kΩ
R3	8.2MΩ
R4	4.7kΩ
R5	1.8kΩ
R6	10kΩ
R7	100kΩ
R8	10kΩ
All 1/4 watt carbon ±5%	

See
**Shop
Talk**
page 281

Potentiometers

VR1	47kΩ linear carbon
VR2	220kΩ linear carbon

Capacitors

C1	1000μF 25V elect.
C2	100nF polyester (C280)
C3	100nF polyester (C280)
C4	100μF 16V elect.
C5	100μF 16V elect.
C6	220nF polyester (C280)
C7	10nF polyester (C280)
C8	22nF polyester (C280)
C9	22nF polyester (C280)

Semiconductors

IC1	7815 (15V 1A regulator)
IC2	555 timer i.c.
IC3	555 timer i.c.
TR1	BC179 <i>pnp</i> silicon
TR2	TIP41A <i>nnp</i> silicon
D1,2,3,4	1N4002 (4 off)
D5,6	1N4001 (2 off)

Miscellaneous

T1	mains transformer having twin 15V 2A secondaries
S1	d.p.d.t. toggle switch
S2	d.p.s.t. toggle switch
LP1	mains neon indicator having integral series resistor
SK1,2,3,4	wander sockets (4 off)
Aluminium case type ABS M6007, 170 x 213 x 82mm (rear) x 31mm (front); 0.1 inch matrix stripboard, 20 holes by 50 strips; two control knobs; two heatsinks; mains lead; standard 3-pin mains plug.	

Approx. cost
Guidance only **£13.00**

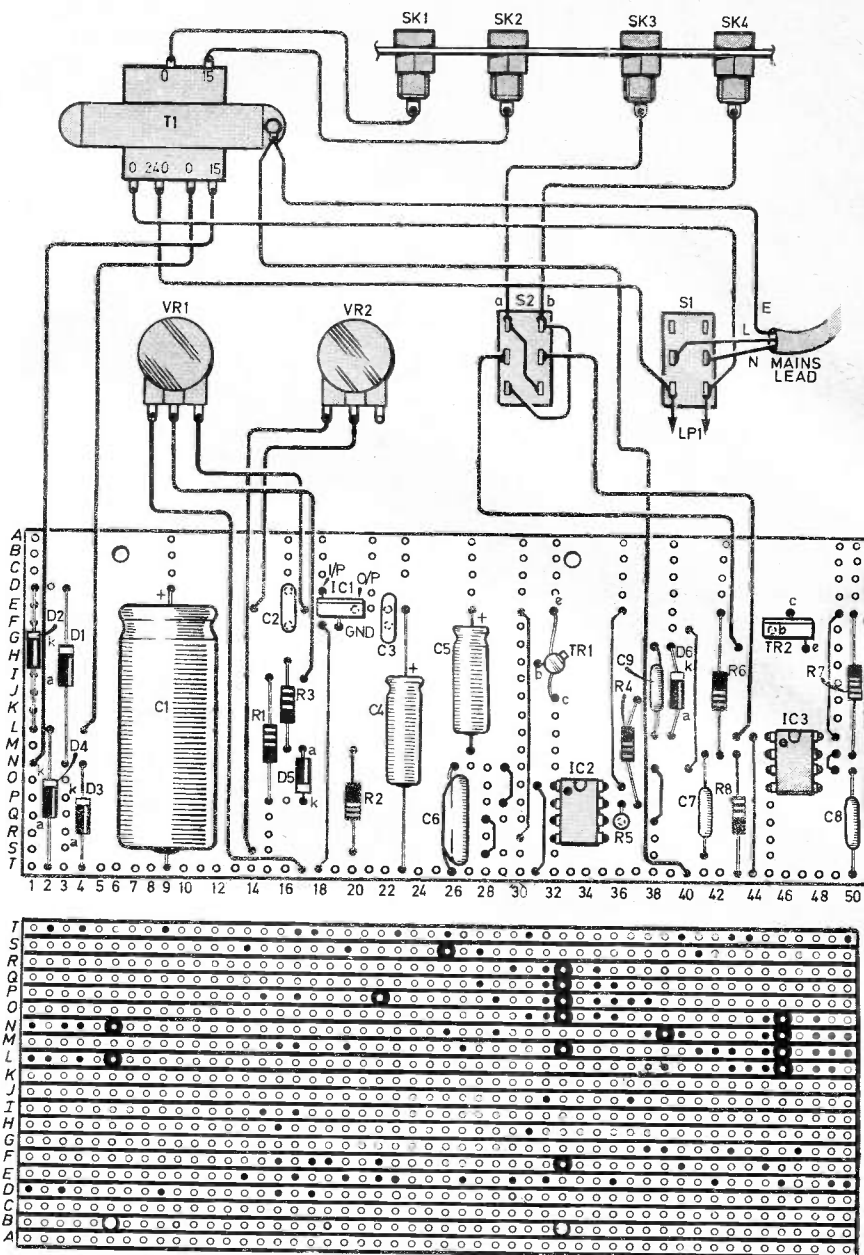
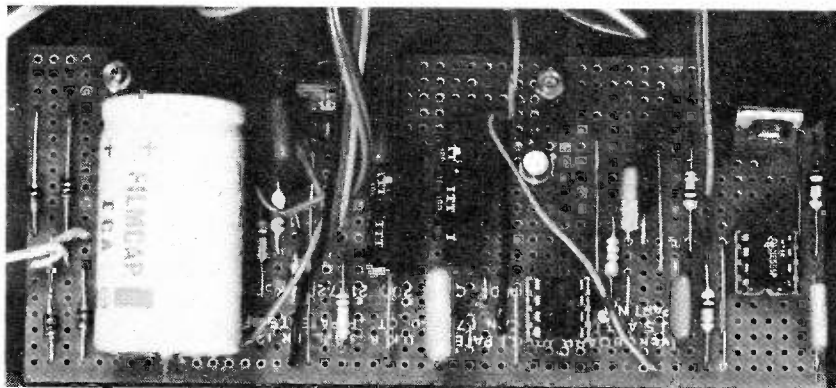


Fig. 3. Full component layout, interwiring and constructional details of the Train Controller.



The prototype component board.

controlled by a voltage applied to the junction of R2, R3. R3 is connected in the base circuit of TR1 to effectively convert TR1 to a voltage rather than current operated device, and it also greatly boosts the input impedance of TR1.

CONTROL VOLTAGE

If the base terminal of TR1 is at or near the positive supply potential, TR1 is cut off and supplies no current to the timing components of the oscillator. This prevents oscillation and gives no output pulses so that there is zero output power from the unit.

Taking the control voltage at the junction of R2 and R3 down towards the negative supply potential takes TR1 into conduction and activates the oscillator. The more negative the control voltage is taken, the heavier TR1 conducts, the higher the operating frequency of IC2, and the greater the output power of the circuit. Full output power is achieved when the control voltage is at nearly the negative supply rail voltage.

VR1 provides the control voltage and it therefore acts as the throttle, but the output from its slider is fed to the v.c.o. via R1 and D5. When the throttle is advanced by taking the slider of VR1 down towards the negative track connection this decrease in control voltage is coupled to R3 by way of R1 and D5, but there is a short delay while C5 charges up. This gives the simulated inertia.

If VR1 has its slider moved back up towards the positive track connection, D5 becomes reverse biased and blocks any significant discharge current through D5, R1 and VR1. A discharge current will flow through R2 and VR2 though, but there will be a substantial decay time if VR2 is at maximum resistance. This gives the simulated momentum.

The discharge time can be greatly reduced by setting VR2 for decreased

resistance, and this produces the simulated braking.

The diode D6 is used to suppress any high voltage spikes which might otherwise be generated across the motor and which could damage the unit. C9 attenuates the high frequency components on the output of the unit, and this reduces the risk of the unit causing radio interference. S2 is a straightforward reversing switch.

COMPONENT BOARD

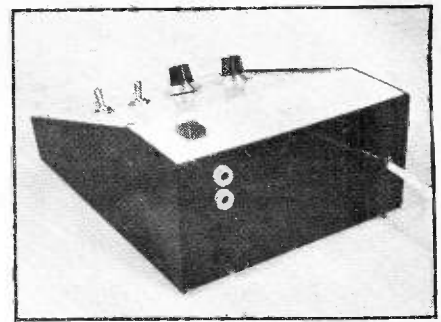
The positioning of the components on the topside of the 0.1in matrix stripboard can be seen in Fig. 3 together with the 15 breaks to be made in the copper strips on the underside of the board. Produce the breaks using either the special spot-face cutting tool or a small twist drill. Drill the two mounting holes for M3 fixings.

Next solder the components and link wires into place leaving the semiconductors until last. Be careful to connect the polarised capacitors and semiconductor devices the right way round.

CASE

A sloping front case is used as the housing for the prototype, however any case having approximate dimensions of 170 x 213 x 82mm will make a suitable housing for the controller. It is not essential to use a sloping front type. The four controls are mounted on the sloping part of the front panel leaving a space for the component board on the base of the case behind the controls.

LP1 is mounted on the horizontal part of the front panel so that there is sufficient space for the mains transformer to the left of this. The mains transformer can be any type capable of delivering 15 volts at 2A twice, and the component used in the prototype is a tapped type with the unwanted tappings being ignored. A solder tag is fitted on one of the fixing bolts of T1.



Rear view of case showing sockets.

The four output sockets are mounted on the rear panel of the case. Wander sockets are used on the prototype, but terminal posts can be used if preferred.

The off-board components are wired to the component panel using p.v.c. insulated multi-strand wire, and the other interconnections are completed in the same way. This wiring is shown in Fig. 4. An entrance hole for the mains lead must be drilled in the rear panel of the case, and the mains earth lead must connect to the solder tag fitted on one of the fixings of T1.

Finally, fit small finned heatsinks to TR2 and IC1 to prevent them from overheating in use, and bolt the component panel in place. It is advisable to place some insulation tape over any exposed mains connections to avoid the possibility of these coming into contact with anything.

It is strongly recommended that the wiring should be checked thoroughly several times before the unit is connected to the mains supply and switched on. T1 is capable of providing quite high currents, and wiring errors or solder blobs causing short circuits between copper strips on the component panel could easily cause damage to the unit. □

CIRCUIT EXCHANGE

ELECTRONIC SWITCH/DOORBELL ALARM

THIS is a very versatile switching circuit whose output could be made to activate a buzzer or other audible device or any suitable relay or a silicon controlled rectifier. (RLA, CSR2 and associated circuitry illustrates the alternative version.) The contact switches shown could be several in a parallel arrangement, to short the circuit on and used as an anti-theft or other alarm system.

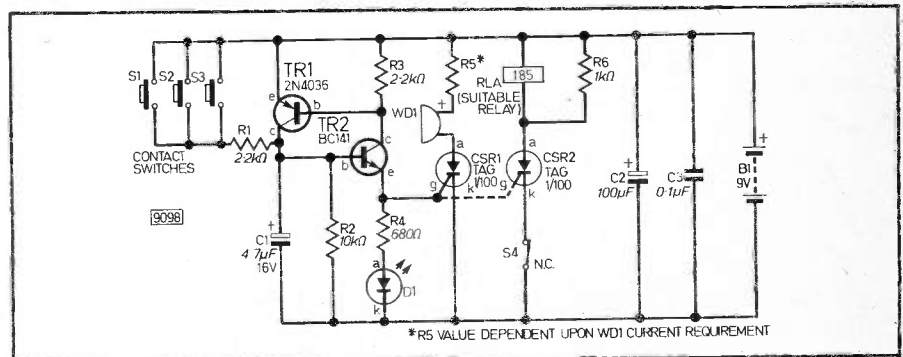
This type of transistor pair switch provides a more definite on-state and

thence overcome false triggering, due to spurious voltages. The 1 kilohm resistor wired in parallel with the relay is to stabilise conduction of the silicon controlled rectifier CSR2 once it is being triggered on, and the normally closed switch S2 is meant to break the circuit, thus switching off the relay. D1 is an l.e.d. added in the emitter line of transistor TR2

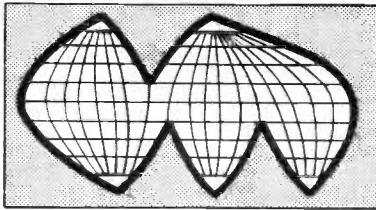
to give on-state indication. (The same circuit could be modified to switch on a 23V a.c. mains line but a triac will have to be substituted for the s.c.r.)

Toh Eng Kiong,
Singapore.

MORE ON PAGE 317



RADIO WORLD



BY PAT HAWKER G3VA

Cost Of Quality

There is no doubt that new technology could give us all superior results to what we are used to with present-day domestic television, radio and audio equipment. The last few years have seen a great revival of development work in the field of high-definition television, in digital systems for radio broadcasting and for audio. The "compact" digital audio disc has reached the UK market—some six years after it was first publicly demonstrated.

Europe seems likely to opt for a fully digital system for some 6 to 8 high-quality sound channels to accompany each video channel on direct-broadcast-satellites (DBS). Indeed, the UK government has decided in favour of the IBA's C-MAC transmission system, one feature of which is a highly flexible all-digital structured system for sound. Initially, digital systems tend to add to the cost of domestic equipment although there are high hopes that costs, particularly of digital memory, will continue to fall.

However, not everybody is seeking higher quality. The video disc, for instance, is capable of providing better quality pictures than most domestic video cassette recorders—though the public appears to consider the "record" ability of tape machines for time-delayed viewing more attractive than the better quality of the video discs.

Laser Discs

For these reasons the whole industry is watching very carefully the public response to the digital audio disc in the UK and the USA. More than 35 manufacturers are licensed to make laser-type players for the 4.7-inch "Compact Disc" format as developed by Sony and Philips.

The ability of the digital disc to provide virtually noise-free reproduction, with almost a (theoretical) 90dB dynamic range (compared with about 60dB for the best analogue recordings), free of wow-and-flutter and with no inherent reason for degradation of quality with frequent playing, represents a tremendous improvement. But, of course, not every existing hi fi installation comes near to providing a distortionless 90dB range which requires an instantaneous peak-power rating of hundreds of watts.

Digits do nothing to overcome deficiencies in loudspeakers or room acoustics, and a system capable of doing real justice to these discs might easily give you more problems with your neighbours than even an a.m. CB transceiver! Then again there will for many years be a need for a conventional disc player if only to play-out existing collections—and recordings that are not yet available in the Compact format.

For all these reasons the electronics consumer industry, the recording industry and the broadcasters will be watching intently to see just how many people invest in the new digital equipment, and whether this looks like rising steadily over the next few years.

Radio Data

The Compact disc results could, for example, influence decisions on the new "Radiodata" system, now close to European standardisation. Radiodata puts digital data signals onto v.h.f./f.m. radio broadcast transmissions which can be dis-

A.M. and CB

Although the Citizen's Band licences introduced by the Home Office in November 1981 were quite specifically for "angle-modulation"—the rather fancy term that covers both frequency modulation (f.m.) and phase-modulation (p.m.)—a considerable number of amplitude-modulated (a.m.) signals can still be heard on 27MHz in the UK. A great deal of these signals on channels outside the official UK band.

While many of the users of a.m. are clearly those who bought equipment before UK licences were issued, there does seem to be a firmly-held view that a.m. is the "better" system and that the Home Office was just intent on making life difficult for the early birds in opting for f.m. One correspondent recently suggested to me that the Home Office had "conned" the public into believing that a.m. caused interference to television sets, whereas f.m. did not.

While I would agree that the whole protracted shambles over the introduction of CB licences in the UK did not reflect much credit on the licensing authorities, there really are solid technical reasons for asserting that f.m. is considerably less likely to cause interference to television, radio and audio domestic equipment than a.m. The fact is that a.m. signals are easily demodulated in any non-linear stage, whereas f.m. requires some form of off-resonant tuned circuit to convert it first into a.m. before it can be demodulated by any diode or other non-linear component.

That is not to say that consumer equipment should not be designed to cope with

played on a suitable receiver in the form of an alpha-numerical display. This provides station identification, accurate time and—if there are sufficient display devices—a short programme-related message.

The data signals can also be used for several forms of automatic tuning or adjustment of controls. For example, a car radio could be made to follow a particular programme channel throughout a long journey.

It is all a technically elegant scheme. But at the moment the necessary alpha-numerical display devices and the associated electronics seem bound to add substantially to the cost of radio receivers even when special-purpose large-scale-integrated circuits become available. Some engineers believe that people would be prepared to pay for these facilities for "top-of-the-range" car models and possibly domestic hi fi tuners; others are not so sure.

The system is the outcome of European co-operation with the preferred system developed in Sweden and closely akin to a BBC-developed system. But its future depends on whether semiconductor manufacturers are willing to take the risk of investing a substantial sum in new LSI chips. Otherwise the system may never "take off".

local low-power a.m. transmissions. This is usually possible without adding greatly to the cost. But the fact is that a great deal of domestic equipment in use today (and for the foreseeable future) cannot cope with local a.m. signals, and some of it does not even cope well with local f.m.

Although the interference statistics are gradually improving, there is no doubt that CB has in some regions overwhelmed the interference investigation teams and must be one of the reasons why British Telecom are hoping to divest themselves of responsibility for this work. The UK is not alone in opting for f.m.-only; this applies also in France, West Germany and Holland.

Teddy on the dole?

New electronics technology is often blamed for the loss of employment opportunities. Regrettably one has to admit that there appears to be a basis of truth in this, even though at one time it was claimed firmly that automation would create more jobs than it replaced.

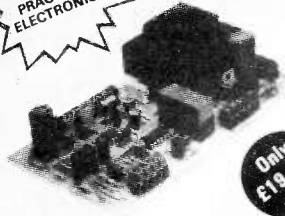
The latest industry that seems to have been hit is traditional toy-making with the American demand slumping as video games take over. Even the video games industry has run into problems as profit margins are slashed and more and more production is being switched away from the USA to the Far East.

The distinction between video games and home computers is also becoming increasingly blurred in a way that never happened with teddy bears and model train sets!

MODULES FOR SECURITY & DETECTION

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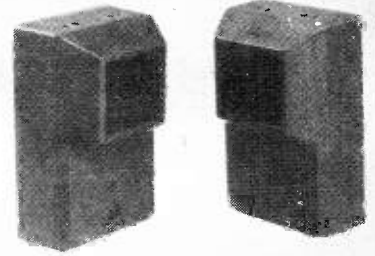
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- Anti-tamper and panic facility
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INFRA-RED SYSTEM IR 1470

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£25.61
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Fully built & tested

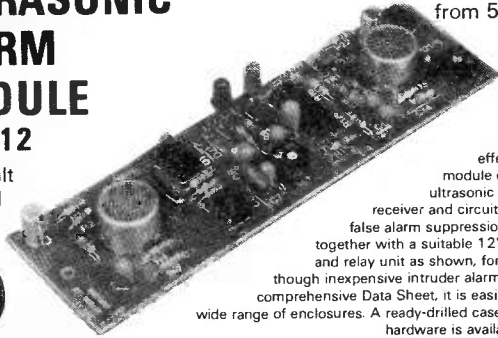
- Range up to 50'
- 12V operation
- Supplied with full instruction
- Easily installed

Now available, a really effective infra-red system built to the high standards demanded by the security industry, and yet offered at this low price. The system consists of a transmitter and receiver which provide an invisible beam over distances from 1-50ft. or more. When the beam is interrupted, a relay is energised in the receiver unit. The use of a modulated beam combined with the infra-red filters, prevent interference from artificial or sunlight, whilst LED indicators ensure easy alignment of the beam. Both units are housed in attractive black moulded enclosures which are easily mounted. Supplied with full instructions, the unit is ideal for use in conjunction with the Control Unit CA 1250 or as an independent unit.

ULTRASONIC ALARM MODULE US 4012

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& tested

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Adjustable range
from 5ft. to 25ft.

A really effective fully built module containing both ultrasonic transmitter and receiver and circuitry for providing false alarm suppression. This module, together with a suitable 12V power supply and relay unit as shown, forms an effective though inexpensive intruder alarm. Supplied with comprehensive Data Sheet, it is easily mounted in a wide range of enclosures. A ready-drilled case and necessary hardware is available (see right).

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TRANSISTOR DATA

BIPOLAR TRANSISTORS

Device	Construction	Case	$I_{C(max)}$ (mA)	V_{CE} (V)	V_{CB} (V)	P_{TOT} (mW)	h_{FE}	f_T (MHz)	Application
AC127	npn Germanium	TO-1	500	12		340	50	2.5	Audio output
AC128	npn Germanium	TO-1	1A	16		700	60-175	1.5	Audio output
AC141	npn Germanium	TO-1	1.2A	18	32	720	40-160	1.5	Audio output
AC142	npn Germanium	TO-1	1.2A	20	32	720	40-160	1.5	Audio output
AC176	npn Germanium	TO-1	1A	18	32	1W	50-250	3	Audio output
AD142	npn Germanium	TO-3	10A	50		30W	30 min	0.5	Audio output
AD149	npn Germanium	TO-3	3.5A	50	50	22W	30-100	0.5	Audio output
AD161	npn Germanium	SO-55	3A	20	32	4W	80-320	3	Audio output
AD162	npn Germanium	SO-55	3A	20	32	4W	80-320	1.5	Audio output
BC107	npn Silicon	TO-18	100	45	50	360	110-450	250	Audio driver
BC108	npn Silicon	TO-18	100	20	30	360	110-800	250	General purpose
BC109	npn Silicon	TO-18	100	20	30	360	200-800	250	Low noise audio
BC142	npn Silicon	TO-39	800	60	80	800	20 min.	80	Audio driver
BC143	npn Silicon	TO-39	800	60	60	800	25 min.	160	Audio driver
BC147	npn Silicon	Lockfit	100	45	—	350	110-450	300	Audio driver
BC148	npn Silicon	Lockfit	100	20	—	350	110-450	300	General purpose
BC149	npn Silicon	Lockfit	100	20	—	350	200-800	300	Low noise audio
BC159	npn Silicon	Lockfit	100	30	30	350	500 typ.	150	Low noise audio
BC182L	npn Silicon	TO-92(1)	200	50	60	350	100-480	150	General purpose
BC183L	npn Silicon	TO-92(1)	200	30	45	300	100-850	280	General purpose
BC184L	npn Silicon	TO-92(1)	200	30	45	300	250 min.	150	General purpose
BC212L	npn Silicon	TO-92(1)	200	50	60	300	60-300	200	General purpose
BC213L	npn Silicon	TO-92(1)	200	30	45	300	80-400	350	General purpose
BC214L	npn Silicon	TO-92(1)	200	30	45	300	140-600	200	General purpose
BC327	npn Silicon	TO-92(2)	500	45	50	625	100-600	260	Driver
BC337	npn Silicon	TO-92(2)	500	45	50	625	100-600	200	Audio driver
BC441	npn Silicon	TO-39	2A	60	75	1W	40-250	50	General purpose
BC461	npn Silicon	TO-39	2A	60	75	1W	40-250	50	General purpose
BC477	npn Silicon	TO-18(1)	150	80	80	360	110-950	150	Audio driver
BC478	npn Silicon	TO-18(1)	150	40	40	360	110-800	150	General purpose
BC479	npn Silicon	TO-18(1)	150	40	40	360	110-800	150	Low noise audio
BC547	npn Silicon	TO-92(2)	200	45	50	500	110-800	300	General purpose
BC548	npn Silicon	TO-92(2)	200	30	30	500	110-800	300	General purpose
BC557	npn Silicon	TO-92(2)	200	45	50	500	75-475	150	General purpose
BC558	npn Silicon	TO-92(2)	200	30	30	500	75-850	150	General purpose
BCY71	npn Silicon	TO-18(1)	200	45	45	360	100-400	200	Low noise, general purpose
BD115	npn Silicon	TO-39	150	180	245	800	60 typ.	145	Audio, high voltage
BD131	npn Silicon	TO-126	3A	45	70	15W	20 min.	60	General purpose, medium power
BD132	npn Silicon	TO-126	3A	45	45	15W	20 min.	60	General purpose, medium power
BD135	npn Silicon	SOT-32	1.5A	45	45	12.5W	40-250	50	Audio driver
BD136	npn Silicon	SOT-32	1.5A	45	45	12.5W	40-250	75	Audio driver
BD237	npn Silicon	TO-126	2A	100	—	25W	25 min.	3	Audio output
BD238	npn Silicon	TO-126	2A	100	—	25W	25 min.	3	Audio output
BD437	npn Silicon	TO-126	4A	45	45	36W	40	3	Power switching
BFR40	npn Silicon	TO-92(1)	200	30	30	500	110-800	100	General purpose
BFR50	npn Silicon	TO-92(3)	1A	60	80	1W	15 typ.	50	—
BFR51	npn Silicon	TO-92(3)	1A	40	60	1W	15 typ.	50	—
BFR80	npn Silicon	TO-92(1)	100	50	70	800	100	100	High voltage, general purpose
BFR87	npn Silicon	TO-92(1)	200	160	160	800	30	130	High voltage, general purpose
BFY50	npn Silicon	TO-39	1A	35	80	800	30	60	High voltage, general purpose
BFY51	npn Silicon	TO-39	1A	30	60	800	40	50	General purpose
BFY52	npn Silicon	TO-39	1A	20	40	800	60	50	General purpose
BFX85	npn Silicon	TO-39	1A	60	100	800	70 min.	50	General purpose
BFX88	npn Silicon	TO-39	600	40	40	800	40 min.	100	General purpose
MJE340	npn Silicon	77-03	500	300	300	20W	30 min.	15	High voltage, medium power
MJE520	npn Silicon	77-03	3A	30	30	25W	25 min.	3	General purpose
TIP31A	npn Silicon	TO-220	3A	60	60	40W	10-50	3	Audio output
TIP31C	npn Silicon	TO-220	3A	100	100	40W	10-50	3	Audio output
TIP32A	npn Silicon	TO-220	3A	60	60	40W	10-25	3	Audio output
TIP32C	npn Silicon	TO-220	3A	100	100	40W	10-25	3	Audio output
TIP33A	npn Silicon	TO-66P	10A	60	60	80W	20-100	3	Audio output
TIP33C	npn Silicon	TO-66P	10A	100	100	8W	20-100	3	Audio output
TIP34A	npn Silicon	TO-66P	10A	60	60	80W	20-100	3	Audio output
TIP34C	npn Silicon	TO-66P	10A	100	100	80W	20-100	3	Audio output
TIP41A	npn Silicon	TO-220	6A	60	60	65W	15-75	3	Audio output
TIP41C	npn Silicon	TO-220	6A	100	100	65W	15-75	3	Audio output
TIP42A	npn Silicon	TO-220	6A	60	60	65W	15-75	3	Audio output
TIP47	npn Silicon	TO-220	1A	250	350	40W	30-150	10	High voltage
TIP49	npn Silicon	TO-220	1A	350	450	40W	30-150	10	High voltage

Device	Construction	Case	$I_{C(max)}$ (A)	V_{CE} (V)	V_{CB} (V)	P_{TOT} (W)	h_{FE} (min)	f_T (MHz)	Application
TIP3055	npn Silicon	TO-3P	15A	60	100	90W	5-30	8	General purpose power
ZTX300	npn Silicon	E-Line	500	25	25	300	50-300	150	General purpose
ZTX500	npn Silicon	E-Line	500	25	25	300	50-300	150	General purpose
2N697	npn Silicon	TO-5	1A	40	60	600	40-120	50	Switching
2N2905	npn Silicon	TO-5	600	40	60	600	100-300	200	Switching
2N2926G	npn Silicon	TO-98	100	25	25	360	235	100	General purpose
2N3053	npn Silicon	TO-39	1A	40	60	800	50-250	100	General purpose
2N3054	npn Silicon	TO-66	4A	55	90	25W	25-100	1	Audio output
2N3055	npn Silicon	TO-3	15A	60	100	115W	20-70	1	High power
2N2222	npn Silicon	TO-18(1)	800	40	75	500	35 min.	250	High speed switching
2N3702	npn Silicon	TO-92(1)	500	25	40	360	60 min.	100	General purpose
2N3703	npn Silicon	TO-92(1)	500	30	50	360	30-150	100	General purpose
2N3704	npn Silicon	TO-92(1)	500	30	50	360	100 min.	100	General purpose
2N3705	npn Silicon	TO-92(1)	500	30	50	360	50-150	100	General purpose
2N3706	npn Silicon	TO-92(1)	500	20	40	360	30-600	100	General purpose
2N3771	npn Silicon	TO-3	30A	40	50	150W	15-60	0.8	High Power
2N3772	npn Silicon	TO-3	20A	60	100	150W	15-60	0.8	High Power
2N3773	npn Silicon	TO-3	16A	140	160	150W	15-60	0.7	High Power
2N3904	npn Silicon	TO-92(4)	200	40	60	310	100-300	300	Switching
2N3906	npn Silicon	TO-92(4)	200	40	40	310	100-300	250	Switching

UNIUNCTION TRANSISTORS

Device	Case	V_{b1b2} (V)	$I_{p(max)}$ (μ A)	I_{pk} (A)	P_{TOT} (mW)	n (min-max)
BRY39	TO-72(4)	70	5	2.5	275	Programmable
TIS43	TO-92(9)	35	5	1.5	300	0.55-0.82
2N1671B	TO-5	65	2	-	450	0.47-0.62
2N2646	TO-18(2)	35	5	2	300	0.56-0.75
2N2647	TO-18(2)	35	2	2	300	0.68-0.82
2N4871	TO-92(8)	35	5	1	300	0.7-0.85

ABBREVIATIONS

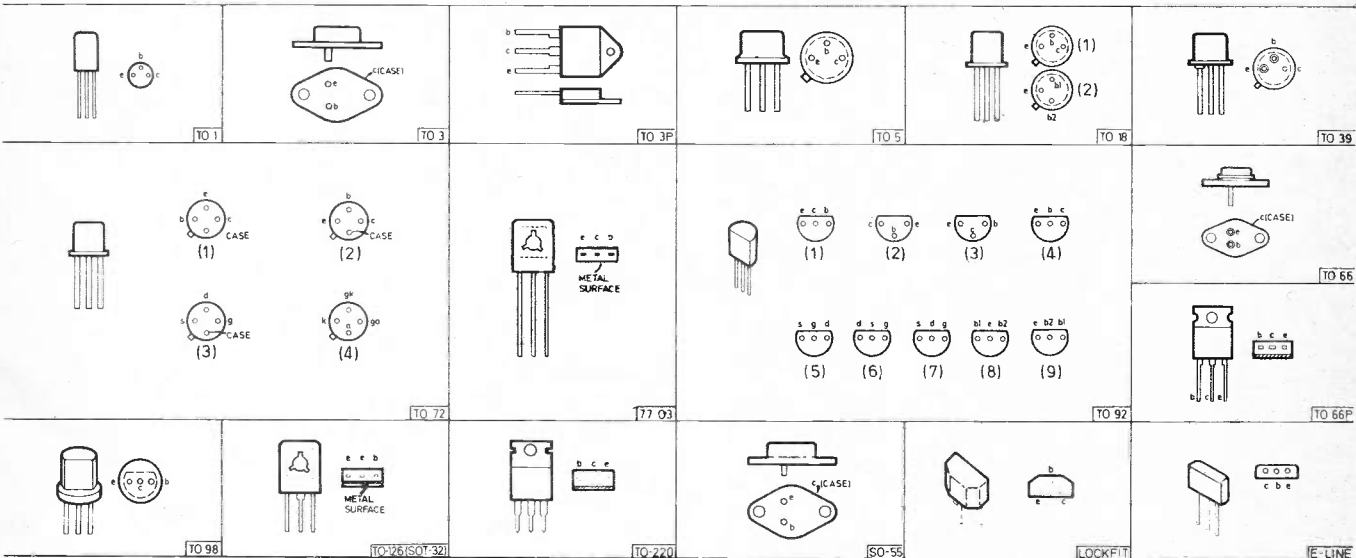
f_T	frequency at which gain reduces to unity
h_{FE}	d.c forward current transfer ratio (gain)
$I_{C(max)}$	maximum continuous collector current
I_{pk}	peak emitter current at avalanche
$ip(max)$	peak point emitter current
P_{TOT}	maximum power dissipation
V_{b1b2}	maximum allowable voltage across b1 and b2
V_{CBO}	maximum collector-to-base voltage, emitter open circuit
V_{CEO}	maximum collector-emitter voltage, base open circuit
n	intrinsic stand-off ratio
npn	} transistor polarity
ppn	

UNITS

V	volts	Ω	ohms
mV	millivolts (10^{-3} V)	k Ω	kilohms ($10^3\Omega$)
μ V	microvolts (10^{-6} V)	M Ω	megohms ($10^6\Omega$)
A	amps	T Ω	teraoohms ($10^{12}\Omega$)
mA	milliamps (10^{-3} A)	kHz	kilohertz (10^3 Hz)
μ A	microamps (10^{-6} A)	MHz	megahertz (10^6 Hz)
nA	nanoamps (10^{-9} A)	μ s	microseconds (10^{-6} s)
W	watts	V/ μ s	volts/microsecond
mW	milliwatts (10^{-3} W)	dB	decibel

TRANSISTOR TYPES 2N4123 AND BCY65EP
For the data on these two free transistors see page 284

TRANSISTOR OUTLINES AND CONNECTION DETAILS



MOISTURE DETECTOR

BY N. KAY

THIS circuit was originally designed to give the alarm when the author's bath water reached a certain depth without having to check the bath every few seconds. However, since completing the unit, it became apparent that it has many other uses, for example, as a rain detector.

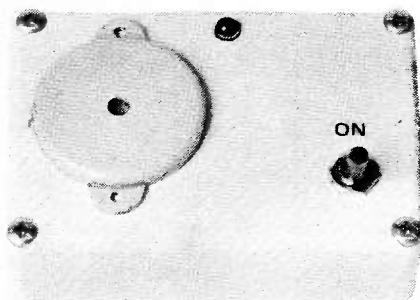
With a suitable sensor, the unit can give the alarm to prevent the weeks washing from getting "re-washed". The sensor could also be fixed in the cold-water tank in the loft to give the alarm should it begin to over fill.

CIRCUIT DESCRIPTION

Many readers may be familiar with the common alarm circuit which use an operational amplifier and detector. Such designs are acceptable but suffer from the drawback that they have a current drain of about 5mA and would quickly flatten a battery and when in constant use, the frequent battery replacement would be costly. The use of a mains powered supply is not recommended in applications involving water, *especially* in a bathroom.

The current consumption of this circuit is extremely low, on one prototype, it was measured to be a fraction of a microamp! The only time when a large current flows (about 8mA) is when the alarm is activated, so one battery should last a very long time.

In the detector circuit (see Fig. 1) an ordinary thyristor CSR1 has its gate terminal connected to the sensor, and R3 is included to limit the gate current and C2 prevents false triggering. The thyristor can be thought of as an electronic switch. When not triggered, it behaves like an open-circuit and no current can flow to the alarm. When triggered the "switch" is closed and the alarm sounds, alerting the user. A visual alarm is given by D1.



The alarm circuit is straightforward, VR1 alters the frequency of the alarm and should be adjusted to give a clearly audible, shrill note. The transformer used in the prototype came from a scrapped radio and almost any miniature audio output transformer with a centre tapped

primary will do. C1 also governs the frequency and by changing its value, different pitched notes can be obtained.

Note that the tab of the thyristor has to be removed in order for it to fit the case. It can be cut with a pair of snips or some pliers, this inconvenience is justified by the fact that this type of device is commonly available from most suppliers.

TRANSDUCER

The Piezo ceramic transducer, WD1 gives a clear audible note and is ideally suited in this application as it is unaffected by steam and moisture and so can be used in the bathroom without damage, unlike an ordinary speaker with its delicate cone. It is not however suitable for use outdoors, and if used as a rain detector, the unit should be kept in the house, with the sensor connected by cable.

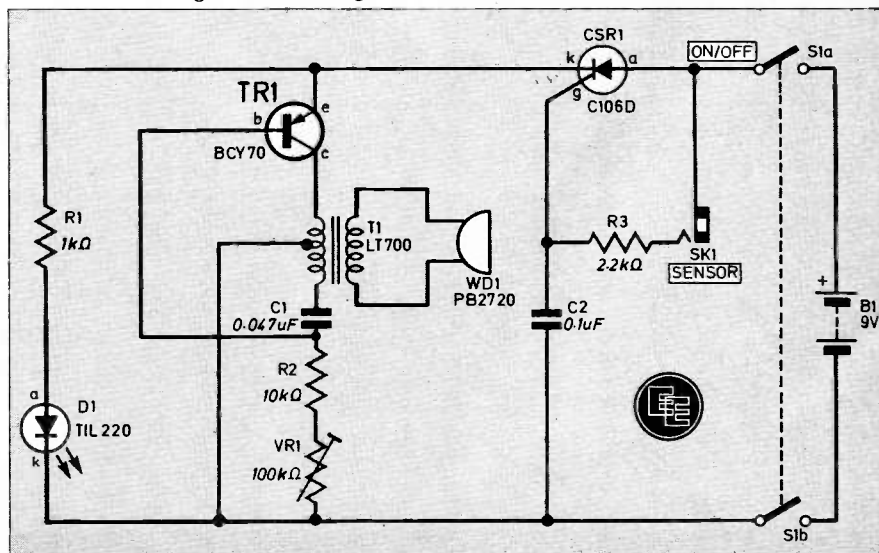
Two distinct advantages are gained by using a thyristor, apart from decreased current drain. Firstly, it is a very sensitive device and will be triggered by the smallest amount of moisture on the sensor and secondly, once triggered, it latches on and so it would, for example, require only one raindrop to bridge the sensor for a continuous alarm to be given.



CIRCUIT BOARD

The circuit (see Fig. 2) is built on a piece of stripboard with a 0.1 inch matrix, 24 holes by 7 strips. The first thing to do after cutting the board to size,

Fig. 1. Circuit diagram of the Moisture Detector.



MOISTURE DETECTOR

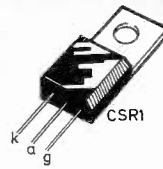
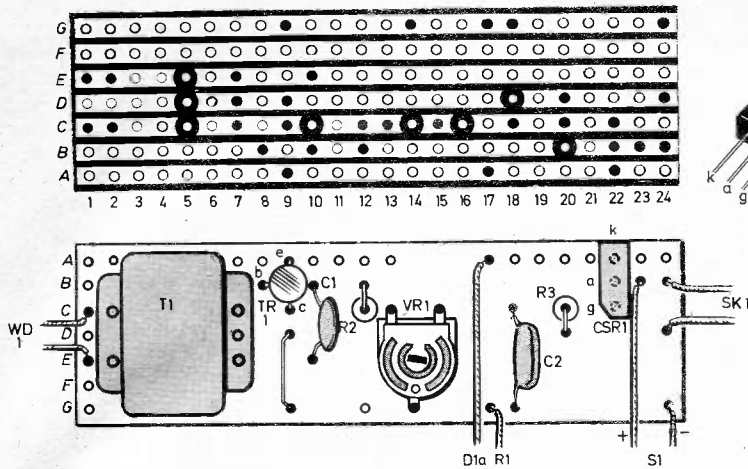


Fig. 2. Layout of the components on the top-side of the stripboard. Underside view shows the positions of the track breaks. Note that this layout is for the LT700 transformer (T1) and other types may have a different pin configuration.

Photograph (right) shows the prototype circuit board. The cropped tag of CSR1 is clearly visible.

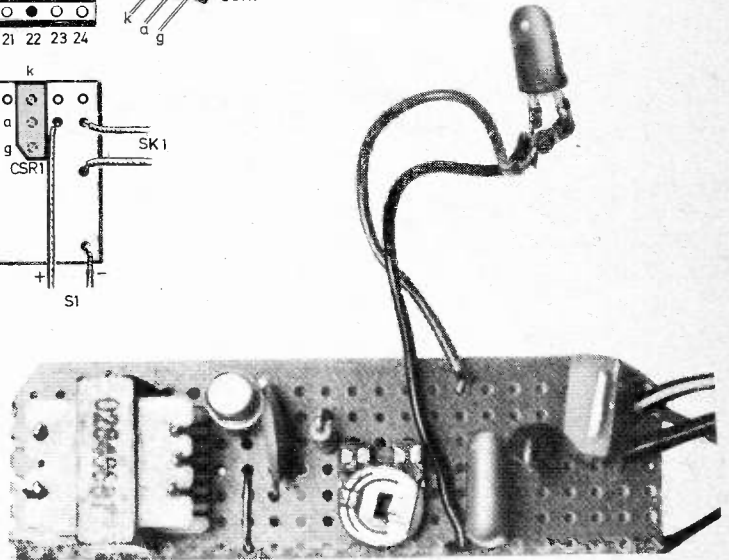
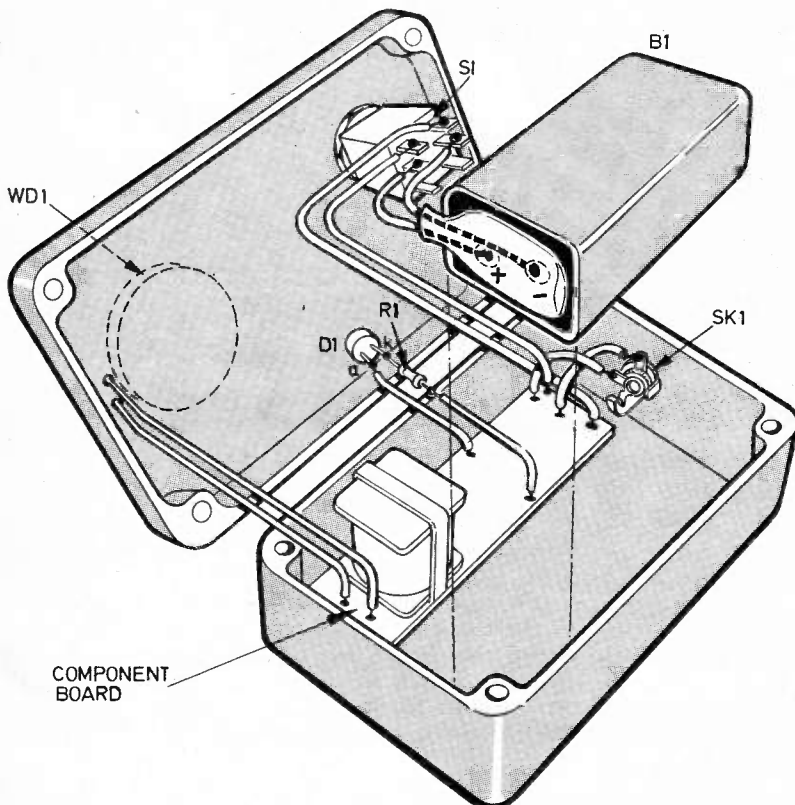


Fig. 3 (below). Final assembly diagram. The transducer is glued to the top of the lid and the wires are fed through two small holes. Note that the cathode (k) of the l.e.d. is signified by a flat on the body.



COMPONENTS

Resistors

- R1 1k Ω
- R2 10k Ω
- R3 2.2k Ω
- All $\frac{1}{4}$ W carbon $\pm 5\%$

Capacitors

- C1 0.047 μ F ceramic or polyester
- C2 0.1 μ F ceramic or polyester

Semiconductors

- D1 TIL220 0.2in. red l.e.d.
- TR1 BCY70 silicon *npn*
- CSR1 C106D thyristor

Miscellaneous

- T1 LT700 audio transformer, 1.2 Ω centre tapped primary, 3.2 Ω secondary
 - WD1 PB2720 piezo ceramic transducer
 - VR1 100k Ω miniature horizontal preset
 - S1 d.p.s.t. sub-miniature toggle
 - SK1 2.5mm jack socket
 - PL1 2.5mm jack plug
 - B1 9V PP3
- Case, 72 x 47 x 25 (Vero type 202-21025); 0.1in. matrix stripboard, 7 strips by 24 holes and 7 strips by 15 holes (for the sensor); battery clip; 7/0.2mm; wire; twin-core cable (approx. 2m); suction cup.

Approx. cost
Guidance only **£6.00**

See
**Shop
Talk**
page 281

is to make the breaks in the tracks (there is a total of 8) and if the special cutter is not available, then a twist drill is ideal (3.5mm is best).

The order in which the components are soldered to the board is not important, however, do not forget to allow for different-sized transformers, depending on the type used.

Once the board is completed, check the wiring, track breaks and the orientation of the thyristor, l.e.d. and transistor. Connect up the battery as shown in Fig. 3 and short the two sensor terminals and the alarm should sound and the l.e.d. illuminate. Once it is certain that the board is correct, the type of sensor required can be decided, according to the unit's intended use.

SENSOR

Fig. 4 shows the sensor designed for the original circuit's use, in the bath. All that is needed is a piece of stripboard, about 7 strips by 15 holes, however any similar size will be suitable. The suction pad was obtained from a suction fixing hook and should be available from most hardware shops, the hook being removed and discarded. Following the diagram, it is a simple matter to construct the sensor.

There should be at least two unused strips between the two sensor contacts, so that they are not bridged by steam or moisture. The contacts should also be tinned to prevent them from tarnishing. In use the sensor should be "stuck" as far away from the taps to avoid splashes.

In the garden, the stripboard sensor would be unsuitable as a thin gap would be needed in order for the contacts to be bridged by a single raindrop and if two adjacent strips were used, there is the possibility they could be bridged by mist or damp. A better idea would be two tinned copper wires placed parallel to each other, separated by a thin gap and fixed a few centimetres above the ground, with no physical base under the wires on which moisture could collect.

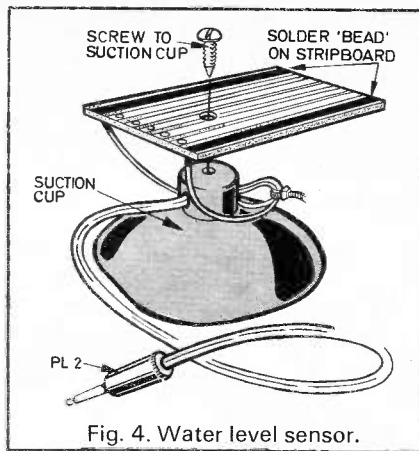


Fig. 4. Water level sensor.

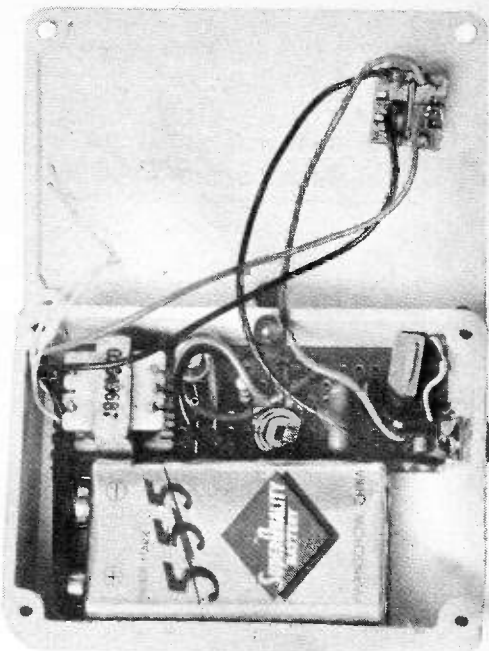
Alternatively, the reader could design and build his own, depending on how it is going to be used.

The sensor was connected to about 2 metres of twin-core cable such as the type used for loudspeaker connections. On the other end, a 2.5mm jack plug was connected, this allows the unit to be used with different type sensors.

When changing the sensors, the unit should be switched off otherwise the alarm will sound each time a jack plug is inserted or withdrawn. To use the bath sensor, it should be affixed to the side of the bath at the opposite end to the taps, the unit should then be switched on. If the alarm immediately sounds, then there is probably some moisture on the sensor. If so, switch off and give it a wipe with a cloth and switch on again. If the unit persists in activating immediately, check to see if the connections to the jack plug and socket are shorting.

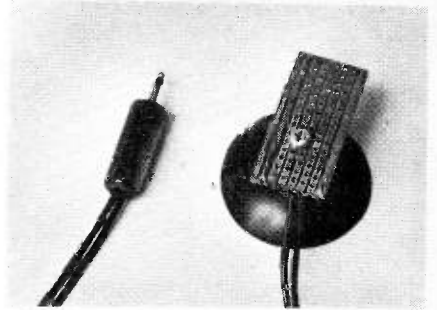
Once the sensor is in place, the unit can be switched on and the bath allowed to fill. If the sensor should become unstuck, the alarm will sound as it will fall into the water!

It may be a good idea to give the unit a test each time, before use, to check that the battery is still in a good condition. □



View inside the finished Moisture Detector unit. The circuit board is held in place with the PP3 battery and the switch is positioned on the lid so as to miss the battery.

The water level sensor made from a piece of stripboard and a suction cup.



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First stage judging has been completed and the 12 finalists have been informed. Models for the second stage judging are to be submitted during June. The Awards for this competition will be made during a special presentation ceremony at Mullard House, London, at the end of June.

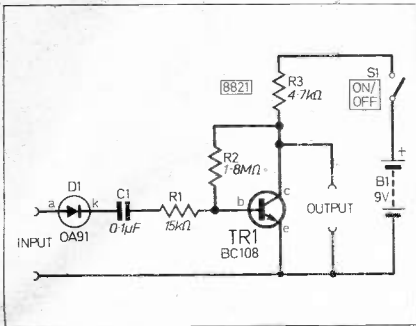
SCHOOLS ELECTRONIC DESIGN AWARD COMPETITION (SEDAC)
SPONSORED BY MULLARD LTD AND EVERYDAY ELECTRONICS

CIRCUIT EXCHANGE

This is the spot where readers pass on to fellow enthusiasts useful and interesting circuits they have themselves devised.

Payment is made for all circuits published in this feature.

Contributions should be accompanied by a letter stating that the circuit idea offered is wholly or in significant part the original work of the sender and that it has not been offered for publication elsewhere.



SOME TIME ago I built an effects unit for use with electrical musical instruments and I thought that I might pass this on. This fuzz box is very simple to construct on stripboard or tagstrip. The circuit relies on the fact that a diode will clip an a.c. signal, but because it crops it, an amplifier is necessary to boost the signal up to a suitable level. The amplifier will run from a PP3 battery and the output can drive either headphones or just go into the input of the power amplifier being used.

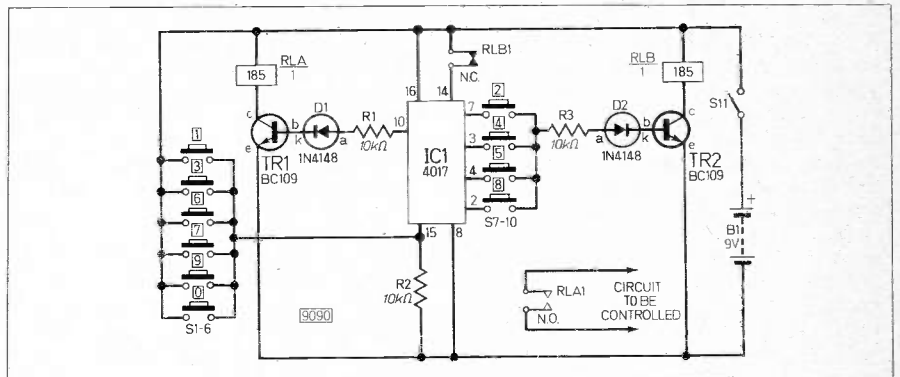
Robert Smith,
Sheffield.

COMBINATION LOCK

WHEN power is applied to the circuit, the first output of the decade counter IC1, pin 3, goes high. If the switch S4 is pressed the base of TR2 goes high, energising the relay RLB. Contacts RLB1 are connected so that when RLB is energised pin 14 goes high, clocking on the counter. The cycle is repeated when switches 8, 5 and 2 are pressed in that order unless switches 1, 3, 6, 7, 9, or 0 are

pressed. If they are, pin 15 goes high, resetting the counter. If the right combination, 4, 8, 5, 2, is pressed pin 10 goes high, energising RLA via TR1, R1 and D1. The contacts RLA1 are used to switch whatever is wanted.

Gideon Tearle,
Headington,
Oxford.



ELECTRONIC KEY

THE Electronic Key in this project is really a 3.5mm jack plug. When it is inserted into a socket an l.e.d., mounted in the jack plug, lights up.

While the plug is left in, the appliance, for example an electronic motor used for opening a door, is switched on. When the door is fully opened a reed switch is operated thus stopping the motor. When the "key" is removed the door automatically closes and another reed switch stops the motor.

The motor will stay in the open state only as long as the plug is inserted, so an "override" switch may be needed to keep the appliance on. The advantage of this system is that the socket can be easily hidden.

Circuit Description

With the jack plug PL1 inserted the l.e.d. is supplied with a voltage to turn it on via the socket SK1 and R1. This also turns on TR1 to operate relay RLA

which has a contact connected in the "motor" supply line.

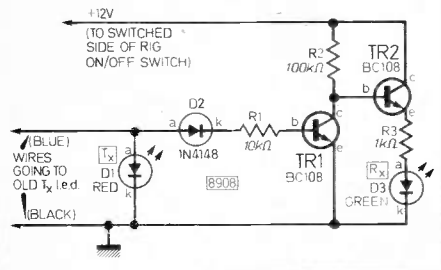
When the plug is removed the contact in SK1 returns to its original position, thus "grounding" the ZTX500 transistor TR2. This operates relay RLB which changes over the polarity of the motor supply, causing the motor to rotate the other way, and the door, which is attached to the motor, to close.

When the door is fully closed the reed switch S2 is opened thus stopping the motor. This leaves the circuit ready for another operation by the insertion of a jack plug.

Another magnet is placed on the door to operate reed switch S1, the "open door" stop switch. The reed switches can be mounted in the door frame and perhaps, for reed switch S1, in a door stop.

Michael Taylor,
Windmillgreen,
Northampton.

CB RIG LIGHT ADD-ON

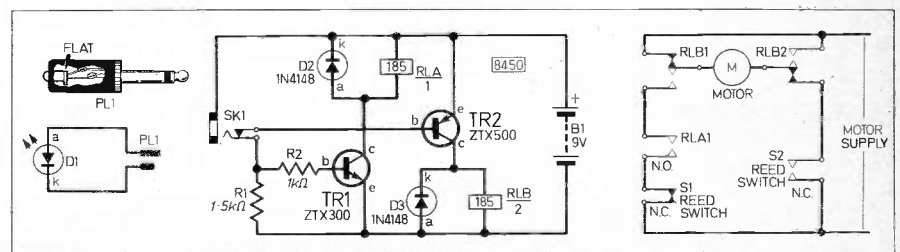


THIS circuit was designed to add a green Rx light to the front panel of a CB rig. I built it into a MAXCOM 4E by removing the small board containing the Tx light and placing the l.e.d. on the new board. I enlarged the screw hole for the Tx board, enough to allow the green l.e.d. through, and drilled a hole through the front panel the same size as the Tx hole.

To fit inside the Rig, the circuit had to be built on a small piece of plain 0.1 inch Veroboard with the l.e.d.s on the other side, and the right distance apart to suit the front panel holes.

Finally, the old Tx wires are connected, then the board covered with insulating tape and then stuck in place.

Michael Palmer,
Havant,
Hants.



SQUARE one FOR BEGINNERS

AFTER the resistor, the capacitor can be thought of as the next most important component in an electronic circuit. Available in a bewildering range of types, sizes and values, the beginner is often confused when it comes to making the right choice for a project.

Basically, a capacitor consists of two metal plates separated by an insulating material known as the **dielectric**. (With the exception of polarised electrolytic capacitors, see *Square One*, last month.) Usually the material used for the dielectric of the capacitor is also used to describe the type of capacitor.

For example, a **polystyrene** capacitor consists of two foil strips (the metal plates) with a layer of polystyrene (the dielectric) between them. This is usually rolled up into a tubular shape with two wire leads, one attached to each plate.

There are many different dielectric materials, including paper, polyester, ceramics and mica to name but a few, and each has different properties making the various capacitors suitable for particular applications. For instance, a silvered mica capacitor can be manufactured to close tolerances and will remain stable and therefore is ideally suited for timing and tuned circuits.

Compare this with the disc ceramic type which has a low tolerance specification, as much as +80% and -20% on the specified value. However, they are available at working voltages of up to 1000V and therefore find their application in power supply spike suppression and decoupling where the actual value is not so important.

The unit of capacitance is the **Farad** but as one farad represents a very high capacity, a more useful unit is the microfarad (μF) which is equal to one millionth ($\times 10^{-6}$) of a farad. Table 1 lists this and the other sub-multiples used for capacitance values.

As with resistors, capacitors are not available in all possible numerical values

TABLE 1. CAPACITANCE SUB-MULTIPLES

picofarads (pF)	= 10^{-12} farads
nanofarads (nF)	= 10^{-9} farads
microfarads (μF)	= 10^{-6} farads
Therefore 1000pF	= 1nF
and 1000nF	= 1 μF
For example 4700pF	= 4.7nF
and 4.7nF	= 0.0047 μF

but in what is known as a **preferred series** of values. The E12 series is most commonly used for capacitors and this consists of the values 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68 and 82 and their decadal multiples and sub-multiples (that is, 1.5, 150, 1500 and so on).

When describing a capacitor, the type of lead formation may also be given and will usually be either **axial** or **radial**. An **axial lead** capacitor will have its leads "sticking" out from each end of the component body (along its *axis*) like a resistor.

The **radial lead** capacitor has both leads *radiating* from one side of its body.

The types of capacitor listed in Table 2 are those most commonly found in the projects featured in **EVERYDAY ELECTRONICS**. All should be available from the regular component suppliers.

TABLE 2. FIXED VALUE CAPACITOR DATA

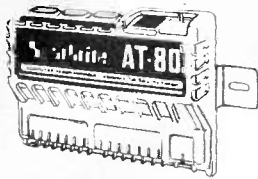
Dielectric type	Capacitance	Tolerance	Working voltage	Typical applications
ALUMINIUM ELECTROLYTIC				
Single ended (can)	100 μF –68,000 μF	+80% –20%	16V–100V	Power supplies, smoothing
Double ended (axial)	1 μF –4700 μF	+50% –10%	10V–450V	
Board mounting (radial)	0.1 μF –2200 μF	+50% –10%	6.3V–63V	
SOLID TANTALUM				Smoothing, decoupling
Miniature bead (radial)	0.1 μF –100 μF	$\pm 20\%$	6.3V–35V	
FOIL AND PAPER				
High capacity (can)	2 μF –10 μF	$\pm 20\%$	600V d.c. (250V a.c.)	High voltage suppression
POLYPROPYLENE				
Epoxy encapsulated (axial)	0.001 μF –0.47 μF	$\pm 20\%$	1000V d.c. (350V a.c.)	Low loss, high voltage
METALLISED POLYCARBONATE				
Brass cased (axial)	1 μF –10 μF	$\pm 5\%$	63V d.c. (45V a.c.)	High stability, professional
Epoxy encapsulated (radial)	0.01 μF –4.7 μF	$\pm 20\%$	630V d.c. (300V a.c.)	
METALLISED POLYESTER				
Epoxy encapsulated	0.01 μF –2.2 μF	$\pm 20\%$	400V d.c. (200V a.c.)	General purpose
Miniature sleeved (radial)	0.001 μF –0.047 μF	$\pm 20\%$	750V d.c. (225V a.c.)	
Dipped case—C280 (radial)	0.01 μF –2.2 μF	$\pm 20\%$ ($\pm 10\%$)	250V d.c. (160V a.c.)	
Moulded case (radial)	0.01 μF –2.2 μF	$\pm 20\%$	400V d.c. (200V a.c.)	
Miniature layer (radial)	0.01 μF –2.2 μF	$\pm 10\%$	400V d.c. (150V a.c.)	
POLYSTYRENE				
Foil tubular (axial)	10pF–0.022 μF	$\pm 2\frac{1}{2}\%$	160V d.c. (40V a.c.)	High stability, filters
SILVERED MICA				
Wax impregnated (radial)	2.2pF–0.01 μF	$\pm 1\%$	350V d.c.	High stability, tuned circuits
CERAMIC				
Monolithic resin dipped (radial)	10pF–1 μF	$\pm 10\%$	100V d.c.	Decoupling
Sub-miniature plate (radial)	2.2pF–220pF	$\pm 2\%$	63V d.c.	Low loss, high stability
Disc (radial)	0.001 μF –0.1 μF	+80% –20%	1000V d.c. (300V a.c.)	Decoupling
Epoxy cased (radial)	10pF–0.1 μF	$\pm 10\%$	100V d.c.	Filtering, coupling

The aluminium electrolytic and solid tantalum capacitors are polarised (that is, they have positive and negative terminals). All others are non-polarised.

Step-by-step fully illustrated assembly and fitting instructions are included together with circuit descriptions. Highest quality components are used throughout.

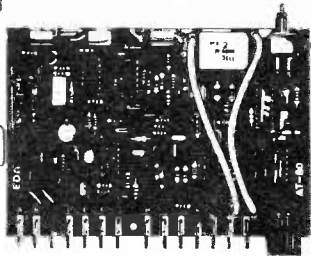
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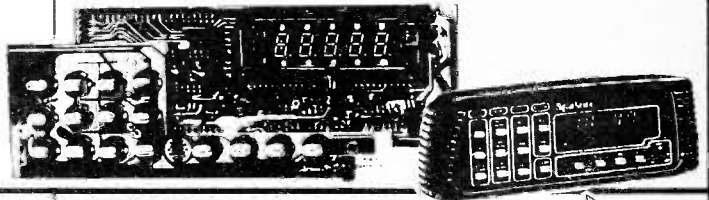
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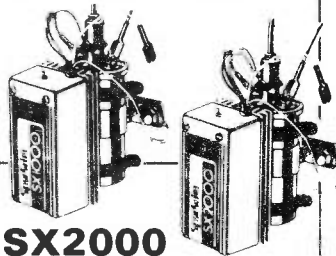
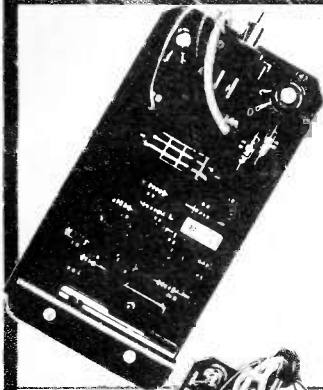
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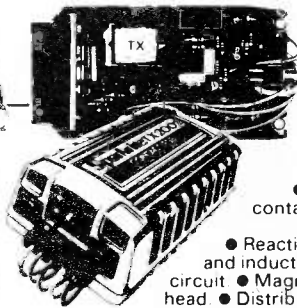


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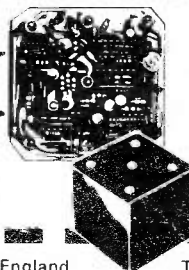


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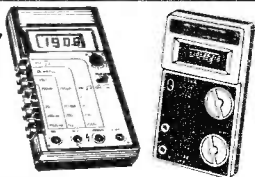
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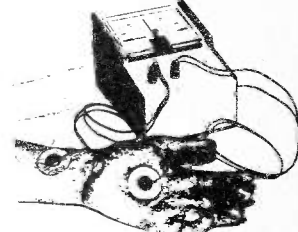
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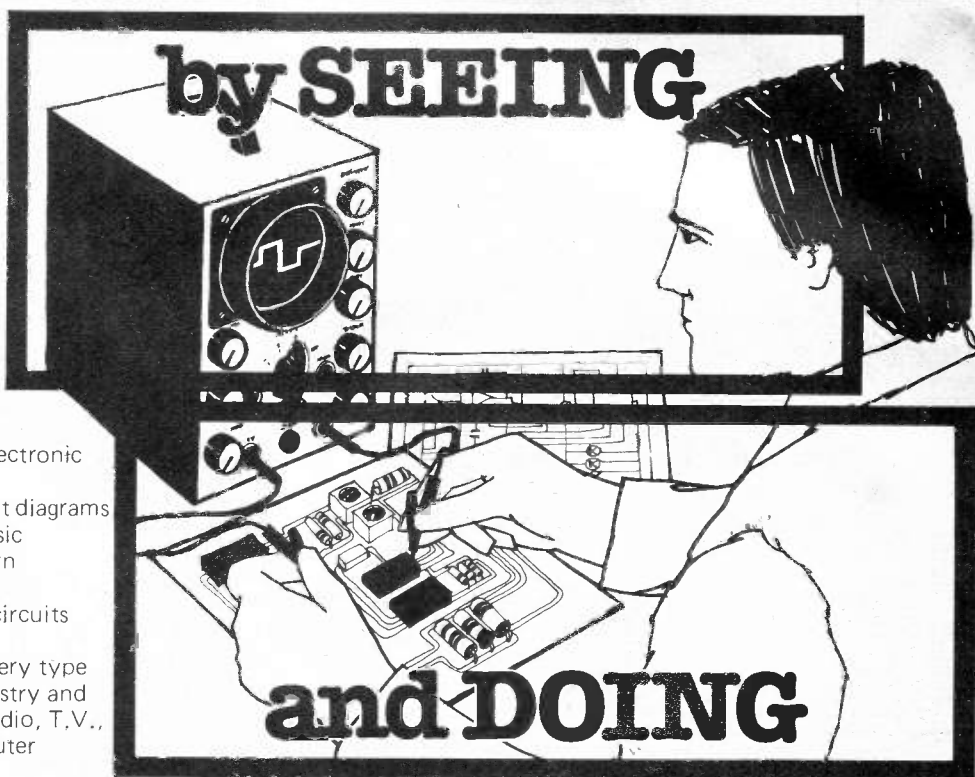
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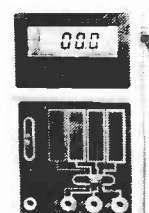
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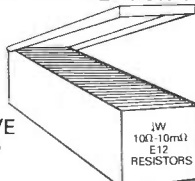
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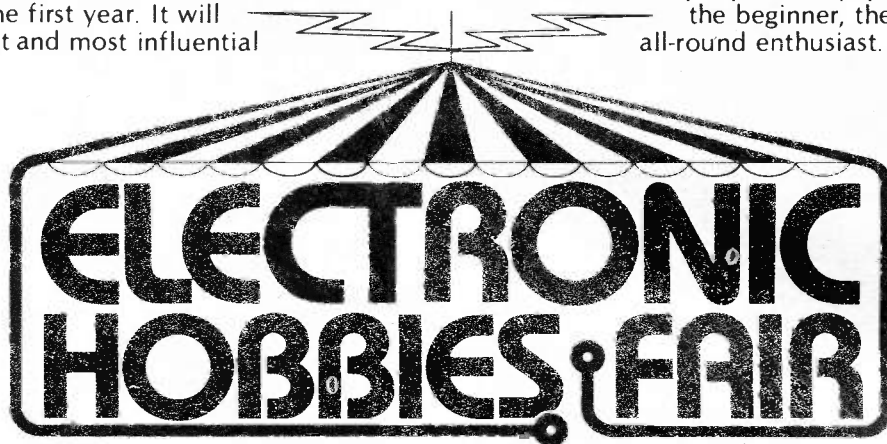
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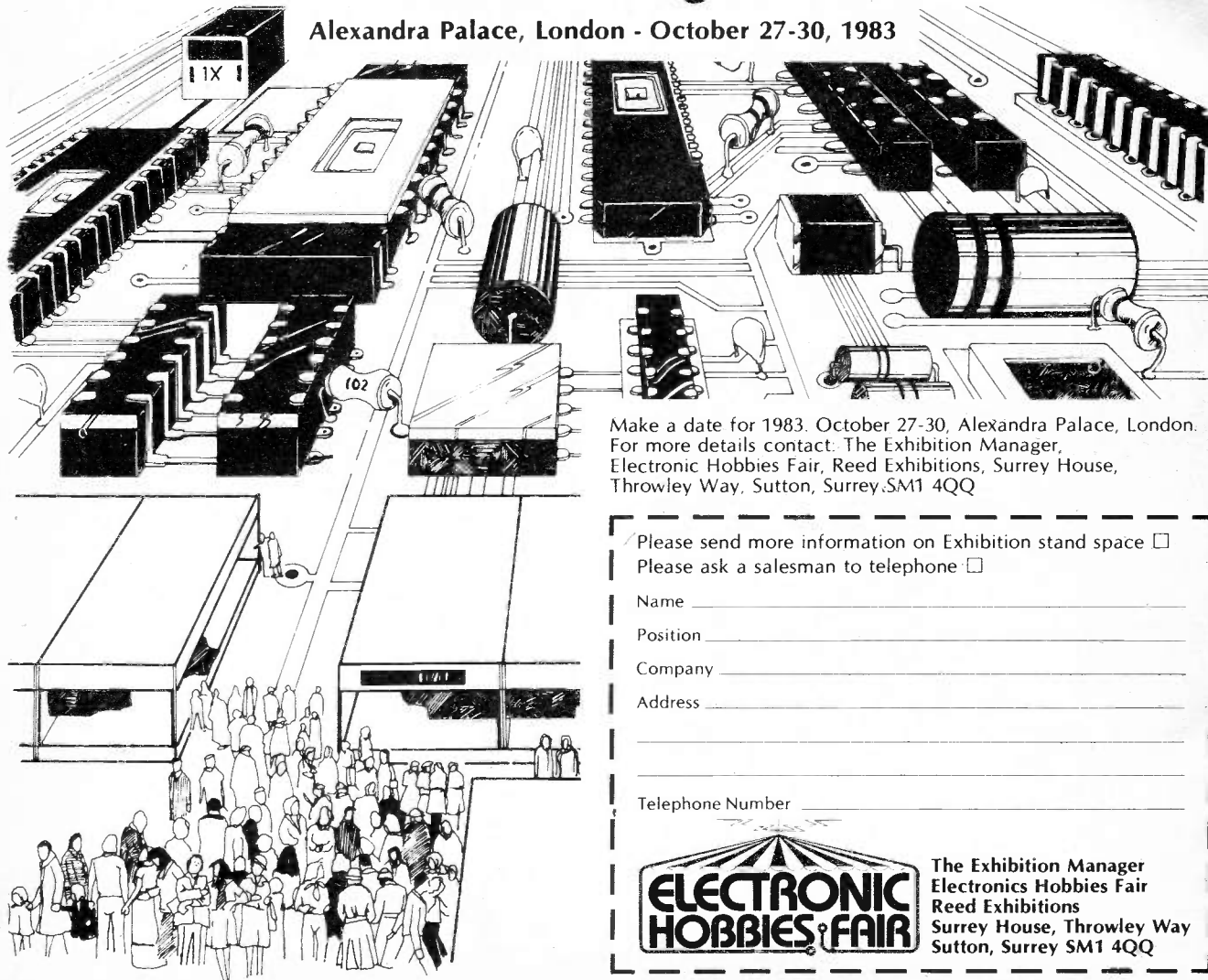
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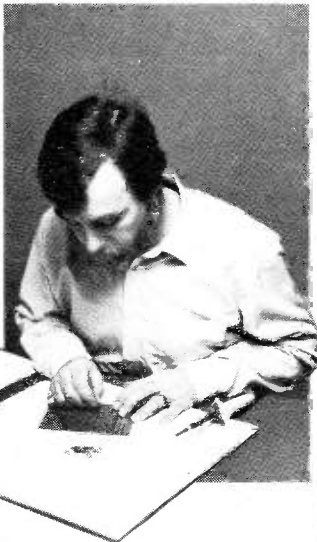
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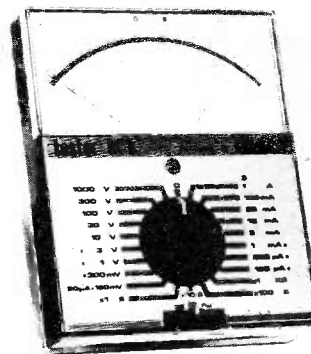
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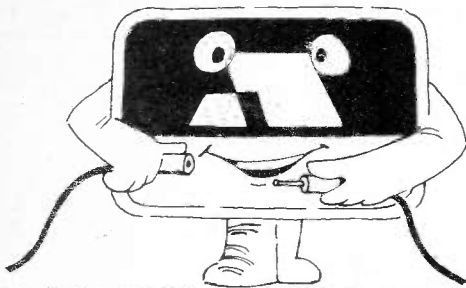
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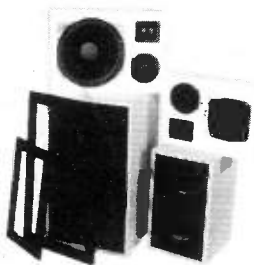
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Price £14.90 each + £2.00 P & P.
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Features: Three digit tape counter. Autostop. Six piano type keys, record, rewind, fast forward, play, stop and eject. Automatic record level control. Main inputs plus secondary inputs for stereo microphones. Input Sensitivity: 100mV to 2V. Input Impedance: 68K. Output level: 400mV to both left and right hand channels. Output Impedance: 10K. Signal to noise ratio: 45dB. Wow and flutter: 0.1%. Power Supply requirements: 18V DC at 300mA. Connections: The left and right hand stereo inputs and outputs are via individual screened leads, all terminated with phono plugs (phono sockets provided). Dimensions: Top panel 5 1/2 x 11 1/2 in. Clearance required under top panel 2 1/2 in. Supplied complete with circuit diagram and connecting diagram. Attractive black and silver finish.
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Supplementary parts for 18V D.C. power supply (transformer, bridge rectifier and smoothing capacitor) £3.50.

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15" 100 watt R.M.S. (HI-FI, P.A., DISCO, BASS GUITAR) Die cast chassis, 2" aluminium voice coil, white cone with aluminium centre dome. 8 ohm imp., Res. Freq. 20Hz., Freq. Resp. to 2.5KHz., Sens. 97dB (As photograph) Price: £32.00 + £3 carriage.

12" 100 watt R.M.S. (HI-FI) Die cast chassis. 2" aluminium voice coil. Black cone. 8 ohm imp., Res. Freq. 20Hz., Freq. Resp. to 4.5KHz. Sens. 95dB. (As photograph) Price: £23.50 + £3 carriage.

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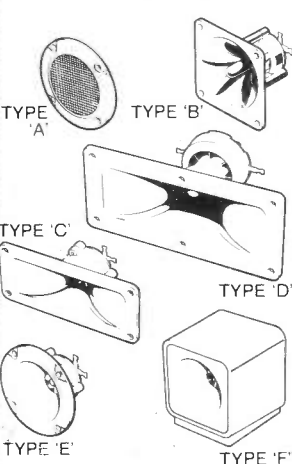
12" 85 watt R.M.S. McKENZIE C1285GP (LEAD GUITAR, KEYBOARD, DISCO) 2" aluminium voice coil, aluminium centre dome, 8 ohm imp., Res. Freq. 45Hz., Freq. Resp. to 6.5KHz., Sens. 98dB. Price: £22.00 + £3 carriage.

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TYPE 'D' (KSN1025A) 2" x 6" wide dispersion horn. Upper frequency response retained extending down to mid range (2KHz). Suitable for high quality Hi-Fi systems and quality discos. Price £7.99 each.

TYPE 'E' (KSN1038A) 3 1/2" horn tweeter with attractive silver finish trim. Suitable for Hi-Fi monitor systems etc. Price £4.99 each.

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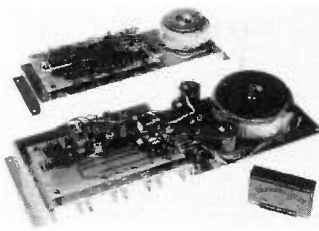
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P256 turntable chassis. ● S shaped tone arm
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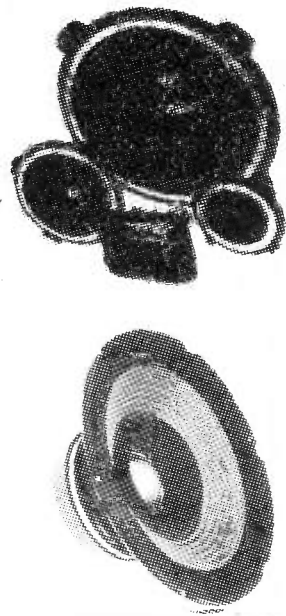
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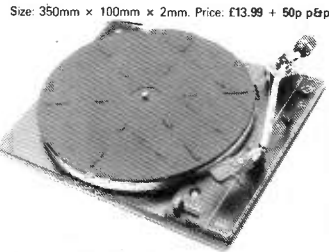
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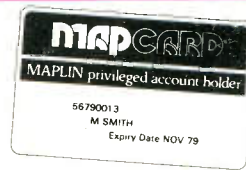
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