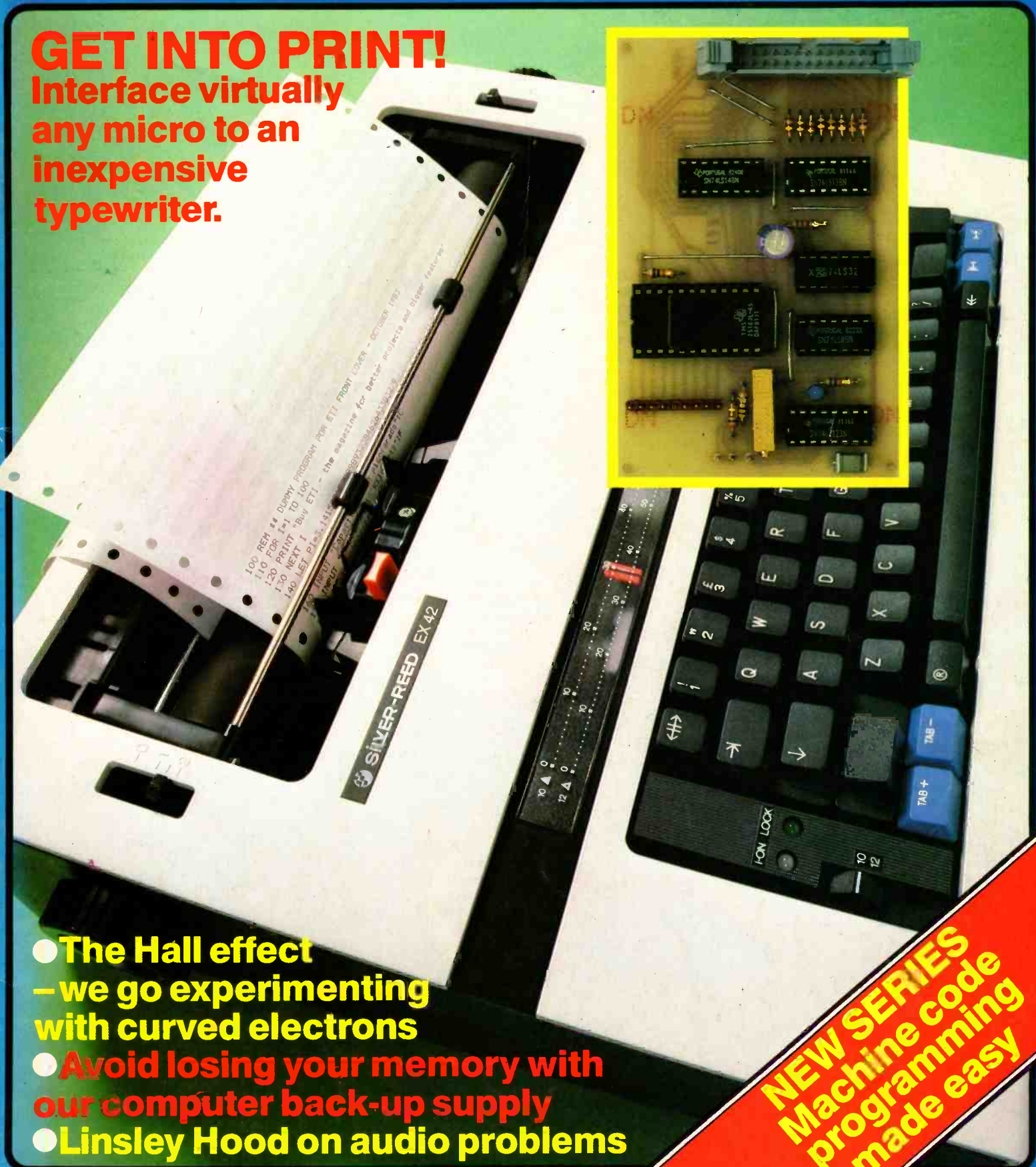


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OCTOBER 1983 85p

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FEATURES

DIGEST13
 News, views and current affairs in electronics.

AUDIO DESIGN.....28
 Continuing this very practical series, John Linsley Hood takes a critical look at the use of ICs in audio, and then goes on to examine audio problems, starting with noise.

MACHINE CODE PROGRAMMING 35
 This new series is our attempt to persuade readers that it isn't as hard as it looks to talk to a microprocessor in its native language. Bob Bennett shows how it' done.

CURVING ELECTRONS.....44
 The Hall effect is a rather useful phenomenon which is the basis of a

fair number of electronic devices. John Dance looks at how these can be employed.

TECH TIPS.....52
 Circuit ideas from you to us and back again.

DON'T SNEEZE.....61
 There are some places on this planet where a sneeze can be very, very costly indeed.

HOW PERMANENT IS PERMANENT?.....71
 Permanent magnetism is used to store some of the greatest pleasures in life (such as music) as well as some of the less pleasant realities (income tax records). But is it reliable?

PROJECTS

TYPEWRITER INTERFACE.....21
 Daisy wheel printers are expensive — daisy wheel typewriters (containing rather more bits) are cheap. (It's called economics.) Here's how to use the latter as the former.

solution to the problem, designed by Phil Walker.

Z80 CONTROLLER COMPUTER..56
 In the concluding part of this project, we give details of the programmer's-eye-view of this system, together with details of some of the useful routines you'll find with it.

ZX80 TAPE MOD.....63
 No, it's not a misprint, we really are telling owners of the '80 how to get reliable SAVE-ing of their programs.

CAR ALARM.....66
 Not another car alarm? No this isn't just another car alarm, but a rather neat design with lots of optional extras — it could be used as a house alarm as well.

• Due to lack of space, IC Update and the Readers' Services Page have been held over



SUPPLY PROTECTOR.....39
 How many times has the flickering of the lights indicated the corrupting of your computer's memory? Here's a

INFORMATION

NEXT MONTH'S ETI..... 9
THIS YEAR'S BREADBOARD EXHIBITION10

ETI BOOKS SERVICE.....69
ETI PCB SERVICE.....77
ADVERTISER' INDEX.....82

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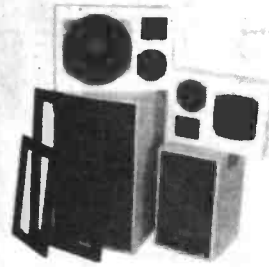
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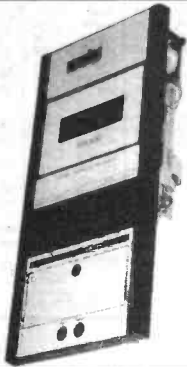
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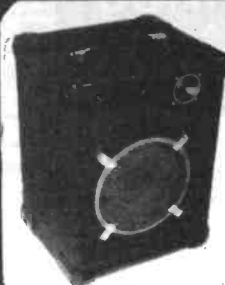
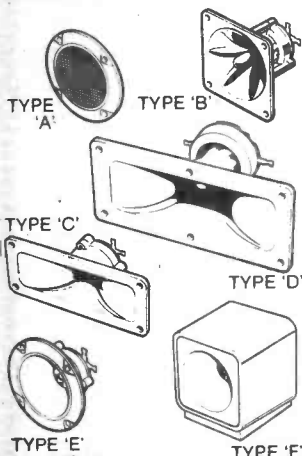
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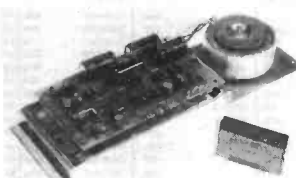
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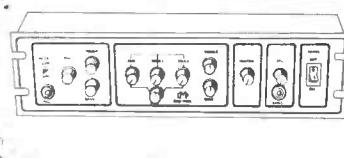
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IDC CONNECTORS

Table of IDC connectors including PCB, Pin, and various connector types with prices.

PANEL METERS

Table of panel meters including FSD, 0-50mA, and other measurement types with prices.

RELAYS

Table of relays including Single Pole, Double Pole, and various ratings with prices.

SLIDE SWITCH

Table of slide switches including DPDT, 1A, and 1P types with prices.

WAFERS

Table of wafer switches including 1 pole, 2 pole, and 3 pole types with prices.

COPPER CLAD BOARDS

Table of copper clad boards including Fibre, Single, and Double types with prices.

DALL SOCKETS

Table of dall sockets including Prof, Low, and Wire types with prices.

EURO CONNECTORS

Table of Euro connectors including Gold Plated, Female, and Male types with prices.

DIY PLUG (Header)

Table of DIY plugs including 14 pin, 16 pin, and 24 pin types with prices.

CRYSTALS

Table of crystals including 32.768kHz, 100kHz, and 1MHz types with prices.

BUZZERS

Table of buzzers including miniature solid-state and PEZO transducers with prices.

ETI PROJECTS

Text advertisement for ETI projects, stating they stock most of the parts.

JUMPER LEADS

Table of jumper leads including single ended DIP and double ended DIP types with prices.

EDGE CONNECTORS

Table of edge connectors including 8 pin, 14 pin, and 24 pin types with prices.

ANTEX SOLDERING IRON

Table of Antex soldering irons including C-15W, G18W, and other models with prices.

25 WAY 'D' CONNECTOR

Table of 25 way 'D' connectors including various pin configurations and prices.

ALUM BOXES

Table of aluminum boxes including 3 x 2 x 1, 4 x 2 x 2, and other sizes with prices.

ASTEC UHF MODULATORS

Table of Astec UHF modulators including Standard 6MHz and Wideband 6MHz types with prices.

SEIKOSHA GP-700

Text advertisement for Seikosha GP-700, a 7 colour graphic printer.

MOS

Table of MOS components including various ICs and their prices.

OPTO ELECTRONICS

Table of optoelectronic components including LEDs, photodiodes, and other devices with prices.

ISOLATORS

Table of isolators including various types and their prices.

COMPUTER CORNER

Text advertisement for Computer Corner, listing various computer hardware products.

FLOPPY DISC DRIVES TEAC

Text advertisement for Teac floppy disc drives, listing various models and prices.

NEC PC8023 PRINTER

Text advertisement for NEC PC8023 printer, highlighting its features and price.

BROTHER 8300 DAISEY WHEEL

Text advertisement for Brother 8300 Daisey Wheel printer, describing its capabilities.

MONITORS

Text advertisement for various monitors, listing models like Microwitec and Kaga.

CASSETTE RECORDERS FOR MICROS

Text advertisement for cassette recorders designed for microcomputers.

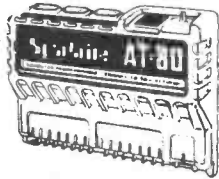
BBC & MICROS PERIPHERALS

Text advertisement for BBC and Micros peripherals, listing various hardware options.

Step-by-step fully illustrated assembly and fitting instructions are included together with circuit descriptions. Highest quality components are used throughout.

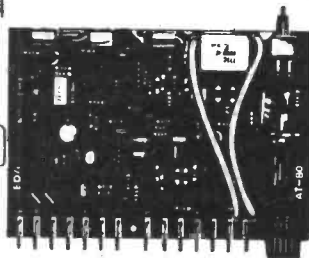
Sparkrite

BRANDLEADING ELECTRONICS
NOW AVAILABLE IN KIT FORM



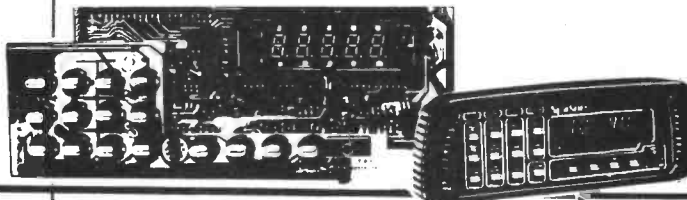
AT-80 Electronic Car Security System

- Arms doors, boot, bonnet and has security loop to protect fog/spot lamps, radio/tape, CB equipment
- Programmable personal code entry system
- Armed and disarmed from outside vehicle using a special magnetic key fob against a windscreen sensor pad adhered to the inside of the screen ● Fits all 12V neg earth vehicles
- Over 250 components to assemble



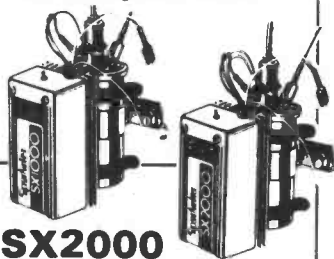
VOYAGER Car Drive Computer

- A most sophisticated accessory ● Utilises a single chip mask programmed microprocessor incorporating a unique programme designed by EDA Sparkrite Ltd. ● Affords 12 functions centred on Fuel, Speed, Distance and Time. ● Visual and Audible alarms warning of Excess Speed, Frost/Ice, Lights-left-on. ● Facility to operate LOG and TRIP functions independently or synchronously. ● Large 10mm high 400ft-L fluorescent display with auto intensity ● Unique speed and fuel transducers giving a programmed accuracy of + or - 1%. ● Large LOG & TRIP memories. 2,000 miles. 180 gallons. 100 hours. ● Full Imperial and Metric calibrations ● Over 300 components to assemble
- A real challenge for the electronics enthusiast!



SX1000 Electronic Ignition

- Inductive Discharge
- Extended coil energy storage circuit
- Contact breaker driven
- Three position changeover switch
- Over 65 components to assemble
- Patented clip-to-coil fitting
- Fits all 12v neg. earth vehicles

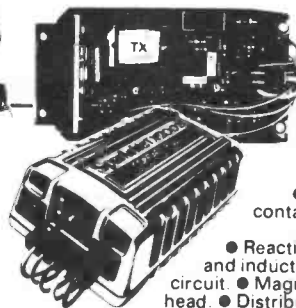


SX2000 Electronic Ignition

- The brandleading system on the market today
- Unique Reactive Discharge
- Combined Inductive and Capacitive Discharge
- Contact breaker driven
- Three position changeover switch
- Over 130 components to assemble
- Patented clip-to-coil fitting
- Fits all 12v neg. earth vehicles

TX1002 Electronic Ignition

- Contactless or contact triggered
- Extended coil energy storage circuit
- Inductive Discharge ● Three position changeover switch ● Distributor triggerhead adaptors included ● Die cast weatherproof case ● Clip-to-coil or remote mounting facility ● Fits majority of 4 & 6 cyl. 12V. neg. earth vehicles ● Over 145 components to assemble.



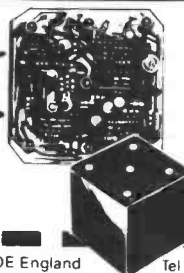
TX2002 Electronic Ignition

- The ultimate system ● Switchable contactless. ● Three position switch with Auxiliary back-up inductive circuit.
- Reactive Discharge Combined capacitive and inductive. ● Extended coil energy storage circuit. ● Magnetic contactless distributor triggerhead. ● Distributor triggerhead adaptors included.
- Can also be triggered by existing contact breakers.
- Die cast waterproof case with clip-to-coil fitting ● Fits majority of 4 and 6 cylinder 12v neg. earth vehicles.
- Over 150 components to assemble

All SPARKRITE products and designs are fully covered by one or more World Patents.

SPECIAL OFFER

"FREE" MAGIDICE KIT WITH ALL ORDERS OVER £45.00



MAGIDICE Electronic Dice

- Not an auto item but great fun for the family
- Total random selection
- Triggered by waving of hand over dice
- Bleeps and flashes during a 4 second tumble sequence
- Throw displayed for 10 seconds
- Auto display of last throw 1 second in 5
- Muting and Off switch on base
- Hours of continuous use from P77 battery
- Over 100 components to assemble

SPARKRITE 82 Bath Street, Walsall, West Midlands, WS1 3DE England

Tel (0922) 614791 Allow 28 days for delivery

	SELF ASSEMBLY KIT
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SX 2000	£19.95
TX 1002	£22.95
TX 2002	£32.95
AT 80	£32.95
VOYAGER	£64.95
MAGIDICE	£9.95

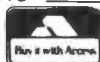
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ADDRESS _____

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£ _____ KIT REF _____

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CUT OUT THE COUPON NOW!

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NEXT
MONTH

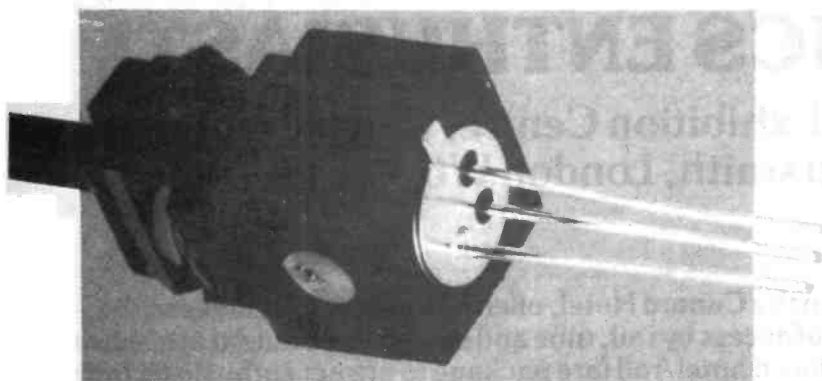
Project Special

As the majority of our readers will be all too aware, the dear old British summer doesn't last that long, and it'll be all too soon before the evenings start to close in, the clocks go back, and thoughts turn back to project work. Well, the November issue of ETI will provide plenty to keep you busy, because in our usual winter effort to lure readers away from the competition (what competition?) we shall be publishing *ten* projects in just the one issue. No silly plastic give-aways from us, we'll just simply deliver where it counts — the contents!

Projects will include a simple design for an active loudspeaker, which, while perhaps not offering the

ultimate in audio performance, will certainly be very cost effective and relatively easy to build; a drum synthesiser module so that you can build up a small kit by using several all set to produce different noises; an add-on unit that will help to make many home-made alarms legal; an analogue amplifier/filter module for use with computer ADCs and DACs; and much, much more that we're still working feverishly on!

Add to this the fact that the next issue of ETI will be over 100 pages in size, and it looks as though there's only one possible choice of magazine for you to buy next month.



Optical Fibres

We've heard all about the industrial and communications use of optical fibres — now ETI takes a look at the home use of them, with, as you would expect from us, a practical guide showing how it's done.



Audio Design

John Linsley Hood continues his examination of some of the problems in audio designing with a look at distortion, and this is followed up by some practical designs for moving magnet and moving coil magnet pick-up amplifiers.

ALL THIS AND MORE IN THE NOVEMBER ISSUE OF ETI, ON SALE OCTOBER 7TH. PLACE YOUR ORDER NOW, OR RISK MISSING OUT!

Articles described here are in an advanced state of preparation. However, circumstances may dictate changes to the final contents.

If you missed
out on Breadboard
'79, '80, '81 and '82.....
Now's the time to catch up with



THE PREMIER SHOW FOR THE ELECTRONICS ENTHUSIAST!

Cunard International Exhibition Centre, Cunard Hotel,
Hammersmith, London W6

Improved Venue

We have transferred BREADBOARD to the Cunard Hotel, offering improved facilities to the visitor, including car parking and ease of access by rail, tube and car, all in a modern attractive setting. See below for our arranged reduced hotel/rail fare package to attract enthusiasts from all parts of the country.

Planned Features include

- Lectures: covering aspects of electronics and computing.
- Electronics/Computing Advice Centre.
- Demonstration: electronic organs/synthesisers.
- Holography presentation.
- Practical demonstration: 'How to produce printed circuit boards'.
- Computer Corner - 'Try before you buy'.
- Amateur Radio Action Centre.
- Computer controlled model railway competition.
- Pick of the projects - Demonstration of the best from ELECTRONICS TODAY INTERNATIONAL, HOBBY ELECTRONICS and ELECTRONICS DIGEST.
- Giant TV screen video games.
- Robotic display.

FRIDAY November 25th

10am - 6pm

SATURDAY November 26th

10am - 6pm

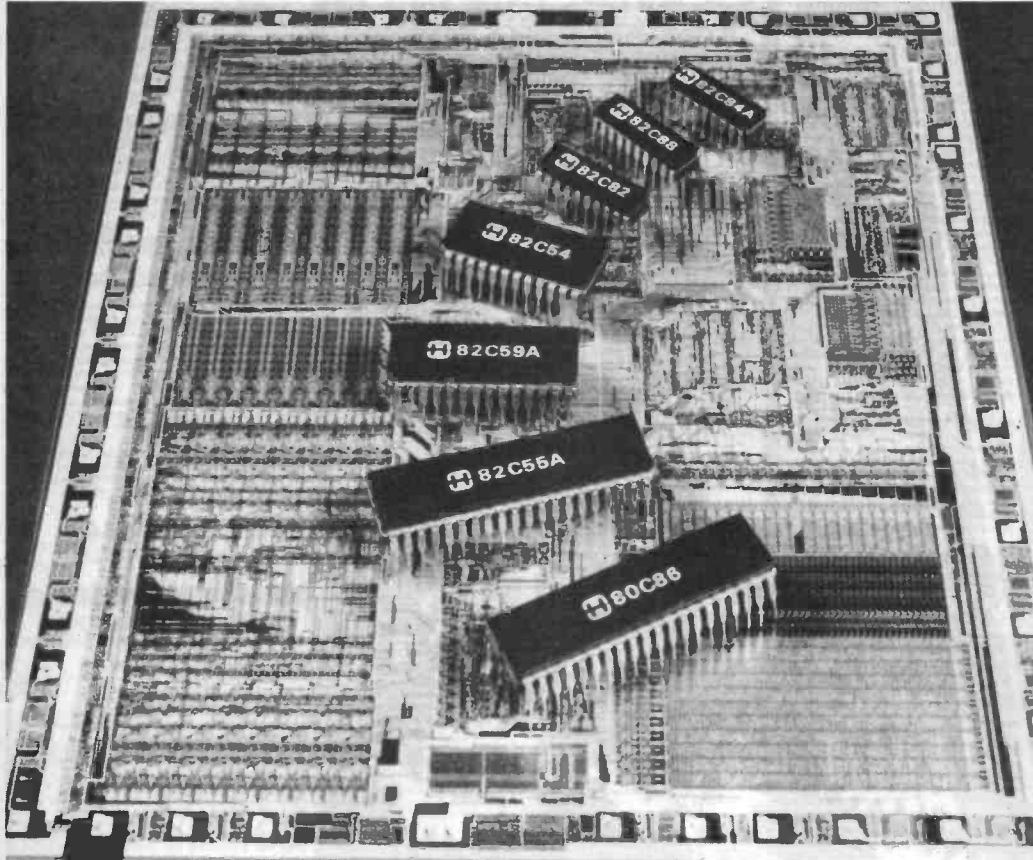
SUNDAY November 27th

10am - 4pm

Why not make your visit to Breadboard '83 the most enjoyable yet. At these *prices, why not bring a partner, do your Christmas shopping, have an evening on the town, see the lights, **and** have a weekend in London!

*see next page

DIGEST



16-Bit CMOS Micro

Harris Semiconductor has announced its 80C86 high performance CMOS 16-bit microprocessor and a family of peripheral devices. Intel Corporation, developer of the industry standard HMOS 8086, will also manufacture the new CMOS devices.

The Harris 80C86 is a completely compatible CMOS alternative for the Intel HMOS 8086. This includes pin-for-pin replacement, TTL level input/output specifications, performance parity at a 5 MHz system frequency

and complete software compatibility. The Harris CMOS 80C86 will operate with existing assembly and high level language programs.

The 80C86 features a standby power supply current of 500 microamperes maximum guaranteed over the full operating temperature and voltage ranges. The operating current of 10 mA/MHz maximum is dramatically lower than the HMOS 8086 which operates at 340 mA maximum at 5 MHz. With CMOS, operating power is reduced as the system's frequency is reduced. The static design of the 80C86 and all other members of the family allows system design

at lower frequencies (down to DC) to further reduce operating power. For maximum power reduction, the system clock can be stopped with all power requirements falling to the standby level (500 μ A for the 80C86, 10 μ A for other family members).

The Harris 80C86 is available now with a 5 MHz operating frequency. The 80C86-2, an 8 MHz system frequency compatible version, will be introduced later.

The 80C86 is packaged in the industry standard 40 pin 0.6" centre ceramic and plastic dual-in-line (DIP) packages. Harris/MHS Semiconductor, Harris Systems Ltd, 153 Farnham Road, Slough, Berks, tel 0753 34666.

Mr Kit Fix-It

There's a new service for all our not-so-nimble fingered readers, being offered by WEB Logic Systems Ltd. They will, for a fee (nothing in this life is completely free) build or repair your kit for you.

WEB say that they will also consider repairing any magazine project provided that it's built on a proper PCB, as published with

the magazine article. What they suggest doing in either case is ringing them and talking over what's involved, and how much it's likely to cost. Alternatively, if you can't phone, they suggest sending the project in with all the information you have (instruction booklet or full magazine article, etc) and authorisation to debit your credit card by up to so much.

WEB say that they will have a

go at — and, more importantly, they have the equipment for — pretty well anything, because they've been offering a similar service to industrial clients for a while now. They expect to be able to turn most jobs around within five days of receipt of the gear.

Further information can be had from WEB Logic Systems Ltd, Gainsborough House, 15 High Street, Harpenden, Herts AL5 2RT, tel 05827 62119.

Breadboard '83

Breadboard '83 will be held on the 25th, 26th, and 27th November at the Cunard International Hotel, Hammersmith. This should make it a lot easier for many people to get to since Hammersmith has good rail and bus connections and the hotel is within a few hundred yards of two NCP car parks.

Breadboard has always had a friendly atmosphere regardless of where it has been held, and moving to the carpeted splendour of the Cunard Hotel should not only preserve this atmosphere but also ensure less wear on the feet!

The dates for the exhibition have been moved nearer to Christmas in response to popular demand... this way, the exhibitors are happy because the visitors are likely to spend more, and the visitors are better placed to drop hints to their loved ones about what they would like in their Christmas stockings instead of socks, after-shave, or cheap perfume!

At last year's exhibition the lectures proved a popular feature, so much so that at times the lecture room threatened to burst at the seams! This year we will again have a broad range of topics, so check well in advance and make sure you're there early to ensure a place.

Model railway enthusiasts will be interested in our computer controlled layout competition. The rules are very simple, amounting to little more than a restriction on the size of the layout.

Bearing in mind the crowds at last year's Computer Corner, the 'advice'/hands on' area will be larger so as to be able to cope with all those would-be purchasers of computers.

For the benefit of those who live a long way from Hammersmith, we have arranged a special package deal that will include rail travel, first class hotel, breakfast, ticket to Breadboard, and various optional extras (tickets to shows, discount vouchers, etc). You can go to the exhibition while your boy/girl friend, husband/wife or whatever does the Christmas shopping and then meet up afterwards.

Breadboard will continue as an exhibition for the enthusiast (or would-be enthusiast!) in the world of electronics. There will be no three ring circuses, fashion shows or the like, just components, magazines, books, equipment, projects, and all the assorted peripherals in which you the reader (enthusiast?) are interested.



Multi-Purpose Scope

Brown Boveri claim to have scored a world first with their M2050 Digital Scope Multimeter. Measuring a mere 257 x 169 x 88 mm this (almost) pocket sized instrument combines a 3½ digit, 32 range digital multimeter, a digital oscilloscope featuring a 128 x 64 dot matrix flat screen display, and a transient recorder with two independent 512 x 8 bit memories.

The M2050 will operate for up to eight hours on its internal NiCad batteries and when switched off will retain information stored in the transient recorder memory for up to three months. One set of input range switching serves both the multimeter and the oscilloscope and the two outputs are displayed simultaneously. The multimeter features AC, AC/DC, and DC operation, will operate down to 15 Hz, and has a true RMS option available on the AC and AC/DC ranges. The oscilloscope has comprehensive triggering facilities, a choice of eight timebases ranging from

100 uS/div to 6 minutes/div, and a roll facility which allows slowly changing inputs to be displayed without irritating flicker. The transient recorder has a maximum sampling rate of 500 kHz and features a save facility which prevents accidental overwriting of stored data.

The manufacturers claim that combining three instruments in this way reduces costs by eliminating unnecessary duplication of common circuitry and speeds up measurement by removing the need to connect up several (often incompatible!) instruments. Aside from such obvious advantages as saving space in the lab and enabling the service engineer to make more sophisticated on-the-spot measurements, suggested applications include the measurement and recording of such variables as temperature and pressure in the field.

The instrument is available as described for £975, or with the addition of a hard copy output for £1150, from John Minister Instruments, 137-139 Sandgate Road, Folkestone, Kent CT20 2DE, tel 0303 41598.

Desk-Top Cases

A new, lightweight case intended primarily for desk-top applications is now available from West Hyde.

Attractively styled with smooth contours, the Empress case is manufactured from 2mm aluminium and has a black and natural anodised finish. It has a sloping top surface at the front

which places switches, knobs and meters at an ideal angle. The four sizes available all have a common profile, so that two or more can be placed side by side on a worktop to form attractive 'suites'.

The Empress case comes complete with self-adhesive feet and is available ex-stock from West Hyde Developments Ltd., Unit 9 Park Street Industrial Estate, Aylesbury, Bucks HP20 1ET, tel 0296 20441.

Portable Computer

Immediate Business Systems have introduced NOMAD, a portable computer the size of a book yet half the weight and which offers the power of a desktop computer.

Nomad provides up to 256K-bytes of non-volatile bubble memory and has been specifically built for outdoor and harsh environment applications, so it is totally waterproof and shock resistant and can operate within temperature ranges of -30°C to +70°C.

Using Microsoft M-BASIC 80 software, Nomad can be programmed directly on its full alphanumeric keyboard, or can be loaded with software developed on any CP/M-based microcomputer. Nomad has a two-line 80 character display panel, but can also be connected via an inbuilt interface (V24-V28/RS232) to drive a visual display unit or printer. Rechargeable batteries make the Nomad totally portable for in the field computing, and the use of a Z80 microprocessor provides fast processing and conservation of power.

Nomad is ideally suited for all applications where portable data processing is required, including such computer-unfriendly environments as oil exploration, steelworks, coal mines, construction, military, etc. Immediate Business Systems plc, 3 Clarendon Drive, Wymbush, Milton Keynes MK8 8DA, tel 0908 568192.

Computer Drive Amplifier

Following the launch of its first computer-drive New Class A amplifier, the SU-V303, Technics have now introduced two further models, the SU-V505 and the SU-V707. Unfortunately, they forgot to send us any bumf on the SU-V707 so we can only tell you about SU-V505!

The SU-V505 incorporates a number of special features. The Computer Drive New Class A system is designed to eliminate transient crossover distortion: sensors feed information on signal level and transistor temperature to a microprocessor which then calculates the optimum bias. To combat electromagnetically induced distortion, all the high current circuitry is placed in what Technics call a *concentrated power block*, effectively isolated from the rest of the amplifier. Other features claimed for the amplifier include the use of linear feedback (no, we don't know what it is either) as opposed to mere negative feedback and a computer controlled protection system. Output is 2 x 60 W into 4 or 8 ohms, both channels driven, with distortion at no more than 0.004% (still rather worse than our own NDFL amp...).

Available in either black or silver finish, the SU-V505 sells for £177.95 and is available through Technics' nationwide dealer network. National Panasonic (UK) Ltd, 300-318 Bath Road, Slough, Berkshire, tel Slough 34522.

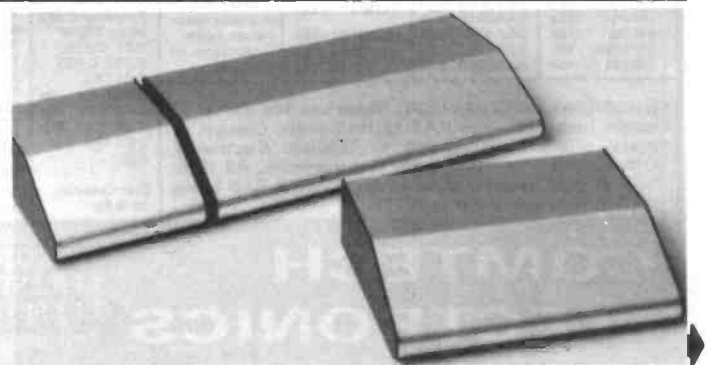
Superbrush

The Superbrush is a compact tool which will clean and polish a variety of surfaces including, but not limited to, metallic materials. Applications include PCB track cleaning, switch and battery contact cleaning, rust removal, etc.

The brush length is controlled by a knurled cap to give a cleaning action ranging from fine

emery cloth to coarse sandpaper. Twist the cap to expose a long brush length for a fine abrasive action, retract the brush until the bristles just protrude for a coarse action.

Recommended retail price of the Superbrush is £2.29 and inquiries are welcomed from stores and distributors as well as end users. Eraser International Ltd, Portway Industrial Estate, Andover, Hants. SP10 3LU.



COMTECH ELECTRONICS

TRANSISTORS				THYRISTORS				CMOS				GPICs				VOLTAGE REGULATORS			
8C1078 12p	8C546 10p	8F195 12p	MPSA 55 20p	2N2907A 25p	C106D 25p	4000 14p	40162 52p	6800 395p	VOLTAGE REGULATORS				Positive						
8C108C 12p	8C546B 9p	8F196 12p	MPSA 56 20p	2N3053 23p	C116D 70p	4001 14p	40163 52p	6802 550p	100mA				78L05 30p						
8C109C 13p	8C547B 9p	8F197 12p	MPSA 63 22p	2N3054 50p	C126D 90p	4002 14p	4501 16p	6808 520p	78L12 30p				78L15 30p						
8C113/4 16p	8C548B 9p	8F198 10p	MPSA 64 22p	2N3055 50p	C126M 98p	4006 50p	4502 55p	6810 250p	78L18 30p				78L24 45p						
8C115/6 17p	8C548C 9p	8F199 12p	MPSA 92 24p	2N3055H 50p	MCR 101 30p	4007 16p	4503 34p	6820 250p	78L24 45p				78L24 45p						
8C119 28p	8C549C 9p	8F200 40p	MPSA 93 24p	2N3440 58p	MCR 102 34p	4008 50p	4504 98p	6821 190p	78L30 30p				78L30 30p						
8C139 32p	8C550C 10p	8F224 15p	TIP 29A 30p	2N3441 120p	T 2800D 130p	4011 14p	4506 70p	6840 595p	78L30 30p				78L30 30p						
8C140 28p	8C556 10p	8F244C 22p	TIP 29B 33p	2N3442 120p	T 2800M 190p	4012 15p	4508 145p	6850 180p	78L30 30p				78L30 30p						
8C141 29p	8C556B 10p	8F245 25p	TIP 29C 33p		2N4443 120p	4013 24p	4510 50p	6852 360p	78L30 30p				78L30 30p						
8C142 27p	8C557B 9p	8F256C 32p	TIP 30A 30p		2N4444 130p	4014 54p	4511 52p	6880 140p	78L30 30p				78L30 30p						
8C143 27p	8C557C 9p	8F256LC 35p	TIP 30B 33p		2N5060 23p	4015 46p	4512 48p	6885 130p	78L30 30p				78L30 30p						
8C160 30p	8C558B 9p	8F257 32p	TIP 30C 36p		2N5061 24p	4016 24p	4513 98p	6889 140p	78L30 30p				78L30 30p						
8C161 32p	8C558C 9p	8F258 32p	TIP 31A 33p		2N5062 29p	4017 42p	4514 110p	68000 POA	78L30 30p				78L30 30p						
8C169C 10p	8C559C 9p	8F259 35p	TIP 31B 35p		2N5064 32p	4018 47p	4515 55p	8726A 140p	78L30 30p				78L30 30p						
8C171B 9p	8C560C 10p	8F236 36p	TIP 31C 37p			4019 30p	4517 180p	8728 120p	78L30 30p				78L30 30p						
8C172C 9p	8C637 20p	8F337 39p	TIP 32A 33p			4020 46p	4518 52p	8795 140p	78L30 30p				78L30 30p						
8C173C 9p	8C638 20p	8F338 40p	TIP 32B 35p			4021 42p	4520 55p	8797 140p	78L30 30p				78L30 30p						
8C177B 17p	8C639 22p	8F457 32p	TIP 32C 37p			4022 46p	4522 65p	8798 140p	78L30 30p				78L30 30p						
8C178C 17p	8C640 22p	8F458 32p	TIP 33A 55p			4023 14p	4526 64p	8080A 400p	78L30 30p				78L30 30p						
8C179 18p	8C70 18p	8F459 37p	TIP 34A 60p			4024 38p	4528 56p	8085A 500p	78L30 30p				78L30 30p						
8C182 9p	8C71 18p	8F494 12p	TIP 41A 46p			4025 14p	4529 72p	8156C 600p	78L30 30p				78L30 30p						
8C182B 9p	8C72 17p	8F595 16p	TIP 41C 60p			4026 84p	4530 92p	8212C 198p	78L30 30p				78L30 30p						
8C183B 9p	8D115 50p	8F596 20p	TIP 42A 60p			4027 29p	4531 70p	8216C 140p	78L30 30p				78L30 30p						
8C183C 9p	8D131 45p	8F597 28p	TIP 42C 60p			4028 42p	4532 88p	8224C 220p	78L30 30p				78L30 30p						
8C184B 9p	8D132 48p	8F598 28p	TIP 110 45p			4029 48p	4541 70p	8228C 450p	78L30 30p				78L30 30p						
8C184C 9p	8D133 70p	8F599 28p	TIP 115 45p			4030 18p	4544 120p	8251AC 495p	78L30 30p				78L30 30p						
8C212 9p	8D135 30p	8F85 28p	TIP 120 70p			4035 56p	4551 88p	8253AC 800p	78L30 30p				78L30 30p						
8C212B 9p	8D136 30p	8F87 27p	TIP 121 70p			4040 48p	4553 180p	8257C 750p	78L30 30p				78L30 30p						
8C213B 9p	8D137 30p	8F88 23p	TIP 122 70p			4041 48p	4554 180p	8259C 900p	78L30 30p				78L30 30p						
8C213C 9p	8D138 35p	8F90 23p	TIP 126 70p			4042 42p	4555 48p	8288 £16	78L30 30p				78L30 30p						
8C214B 9p	8D139 35p	8F91 22p	TIP 127 70p			4043 44p	4556 48p	TMS9980A£22	78L30 30p				78L30 30p						
8C214C 9p	8D140 35p	8F92 22p	TIP 2955 70p			4044 44p	4557 160p	TMS9981N£25	78L30 30p				78L30 30p						
8C237 7p	8D203 70p	8F95A 23p	TIP 3055 70p			4046 55p	4558 110p	TMS9901N590p	78L30 30p				78L30 30p						
8C238 9p	8D204 70p	8U205 140p	TIS44 20p			4048 30p	4559 390p	TMS9902N550p	78L30 30p				78L30 30p						
8C239 9p	8D205 70p	8U206 150p	TIS90 24p			4049 25p	4560 170p	Z80A 4M 580p	78L30 30p				78L30 30p						
8C251 9p	8D206 70p	8U208 140p	TIS92 20p			4050 25p	4566 120p	Z80A P10500p	78L30 30p				78L30 30p						
8C300 40p	8D239A 40p	MJ2500 230p	2N1613 30p			4051 52p	4568 270p	Z80ADMA£16	78L30 30p				78L30 30p						
8C301 35p	8D239C 50p	MJ2501 245p	2N1711 30p			4052 52p	4569 140p	2101.4A 350p	78L30 30p				78L30 30p						
8C302 35p	8D240A 42p	MJ2955 90p	2N1893 30p			4053 54p	4572 42p	2114.4 120p	78L30 30p				78L30 30p						
8C303 35p	8D240C 50p	MJ3000 210p	2N2218 24p			4056 28p	4573 198p	2147.05N600p	78L30 30p				78L30 30p						
8C304 35p	8D241A 42p	MJ3001 225p	2N2218A 25p			4068 15p	4574 198p	2147.45N570p	78L30 30p				78L30 30p						
8C307 10p	8D241C 54p	MJE340 48p	2N2219 24p			4069 15p	4575 198p	2716 400p	78L30 30p				78L30 30p						
8C308 10p	8D242A 42p	MJE350 70p	2N2219A 25p			4070 15p	4580 280p	4164-150N£10	78L30 30p				78L30 30p						
8C309 10p	8D242C 54p	MJE370 80p	2N2221A 20p			4071 15p	4584 38p	4016-200N£7	78L30 30p				78L30 30p						
8C327 12p	8D243A 55p	MJE371 84p	2N2222 18p			4072 15p	4585 74p	5101LC 340p	78L30 30p				78L30 30p						
8C328 12p	8D243C 68p	MJE520 60p	2N2222A 20p			4073 15p	4599 178p	5101LC-1395p	78L30 30p				78L30 30p						
8C332 12p	8D244A 64p	MJE521 68p	2N2368 23p			4074 15p			78L30 30p				78L30 30p						
8C338 12p	8D244C 78p	MJE2955 80p	2N2369A 15p			4075 15p			78L30 30p				78L30 30p						
8C413C 10p	8F180 30p	MJE3055 68p	2N2484 24p			4077 15p			78L30 30p				78L30 30p						
8C414C 10p	8F181 30p	MPSA05 20p	2N2646 48p			4078 15p			78L30 30p				78L30 30p						
8C415C 10p	8F182 30p	MPSA06 20p	2N2904 22p			4081 15p			78L30 30p				78L30 30p						
8C416C 10p	8F183 30p	MPSA12 22p	2N2904A 23p			4082 15p			78L30 30p				78L30 30p						
8C477 25p	8F184 30p	MPSA13 22p	2N2905 23p			4093 32p			78L30 30p				78L30 30p						
8C478 23p	8F185 30p	MPSA42 23p	2N2906A 22p			4098 78p			78L30 30p				78L30 30p						
8C479 24p	8F194 12p	MPSA43 23p	2N2907 23p			40161 52p			78L30 30p				78L30 30p						

LINEAR ICs				CAPACITORS				DIODES				RESISTORS				HARDWARE																																																						
CA3046 72p	LM565 100p	SAA5041 £25	Electrolytic	AA119 9p	1/2 Watt Carbon	4xTO3 mounting kits 20p	64mm Loudspeakers 8R70p	4xTO66 mounting kits 20p	64R 78p each	300mW rating	5 x TO126 bushes/washers 14p	PP3 battery snaps 6p each	10xTO220 bushes/washers 22p	4xHP7 battery holder 22p	5 x TO3P bushes/washers 14p	4xHP7 battery holder 32p	20mm panel fuseholder 30p	20mm chassis fuseholder 7p	4mm plugs 12p	4mm sockets 20p	4mm insulated terminals 30p	3.5mm jack socket 15p	3.5mm jack plug 15p	Phono sockets 15p	Switches: miniature toggle, SPST 60p	SPDT 68p	DPDT 78p each	Miniature push buttons: 1A/250V push to make 16p, push to break 23p	Midget rotary switches, pcb terminals 2P6W, 4P3W, 3P4W 39p each	Slide switch 1A/250V DPDT (22 x 12 x 8 mm) 16p																																								
CA3065 190p	LM567 150p	SAA5050 £10	Tantalum bead,	BAX13 5p	film 5% E 24	CP100 60 IC sockets, 8, 14, and 16 pin 20 of each	440p	CA3080 72p	LM709 35p	SAA5052 £10	CA3086 56p	LM710 70p	SL490 300p	SN76115 98p	0.47 25 5p	0.1 µF 35 12p	CP101 20 BC182/BC213 transistors, 10 of each	130p	CA309A 170p	LM711 60p	SN7660 90p	SN76115 98p	0.47 35 12p	0.22 35 12p	CP102 20 BC183/BC213 transistors, 10 of each	120p	CA3090AQ 300p	LM713 75p	TBA1205 70p	0.68 35 12p	CP103 20 BC184/BC214 transistors, 10 of each	130p	CA3130E 95p	LM747 50p	TBA520Q200p	2.2 16 5p	0.47 35 12p	CP104 20 BC549C/BC559C transistors, 10 of each	130p	CA3140E 44p	LM748 35p	TBA550 240p	2.2 63 6p	0.68 35 12p	CP105 20 BC550C/BC560C transistors, 10 of each	160p	CA3160E 100p	LM1458 36p	TBA700Q200p	4.7 25 8p	2.2 µF 35 12p	CP106 100 1N914 switching diodes, 75-100V/75mA 240p	CP107 100 1N916 switching diodes, 75-100V/75mA 280p	CP108 100 1N4148 switching diodes, 75-100V/75mA 160p	CP109 500 1N4148 switching diodes, 75-100V/75mA 680p	CP110 30 1N4002 1A/400V rectifier 100p	CP111 10 MC1458 Dual op amps (741 type) 320p	CP112 100 400mW Zeners, 4 of each 2V7 to 33V 450p	CP113 100 1.3W Zeners, 4 of each 4V7 to 51V 850p	CP114 4 LF351 JFET op amps, Low noise 170p	CP115 4 LF353 JFET op amps, dual low noise 300p	CP117 50 mixed LED's 3mm/5mm/rect/red/green/yellow 400p	CP119 100 1N4002 1A/100V rectifiers 270p	CP120 10 C106D 400V/4A thyristors 250p	CP121 10 NE555 dual timer IC's 400p	CP124 100 mixed electrolytic capacitors 100p	CP125 5 LM317T 1A/TO220 variable regulators 440p			
CA3161E 140p	LM3900 47p	TBA625 100p	Value/V price	OA47 9p	series 4.7 R	CP126 100 1N914 switching diodes, 75-100V/75mA 240p	CP127 100 1N4148 switching diodes, 75-100V/75mA 160p	CA3240AE 165p	MC1455 16p	TBA800 70p	4.7 63 6p	2.2 µF 35 12p	0.68 35 12p	2.2 µF 35 12p	0.68 35 12p	CP128 100 1N4148 switching diodes, 75-100V/75mA 160p	CP129 100 1N4148 switching diodes, 75-100V/75mA 160p	CA3240E 110p	MC1458 34p	TBA810A 95p	10 µF 16 5p	3.3 µF 35 18p	1N914 3p	1N916 4p	CP130 100 1N4002 1A/400V rectifier 100p	CP131 100 1N4002 1A/400V rectifier 100p	CA3260E 100p	MC1496 70p	TBA820 75p	10 µF 16 5p	6.8 µF 25 20p	1N4148 3p	1N4149 4p	CP132 100 1N4002 1A/400V rectifier 100p	CP133 100 1N4002 1A/400V rectifier 100p	CA3260E 100p	MC1496 70p	TBA820 75p	10 µF 16 5p	6.8 µF 25 20p	1N4148 3p	1N4149 4p	CP134 100 1N4002 1A/400V rectifier 100p	CP135 100 1N4002 1A/400V rectifier 100p	CA3260E 100p	MC1496 70p	TBA820 75p	10 µF 16 5p	6.8 µF 25 20p	1N4148 3p	1N4149 4p	CP136 100 1N4002 1A/400V rectifier 100p	CP137 100 1N4002 1A/400V rectifier 100p	CA3260E 100p	MC1496 70p	TBA820 75p	10 µF 16 5p	6.8 µF 25 20p	1N4148 3p	1N4149 4p	CP138 100 1N4002 1A/400V rectifier 100p	CP139 100 1N4002 1A/400V rectifier 100p	CA3260E 100p	MC1496 70p	TBA820 75p	10 µF 16 5p	6.8 µF 25 20p	1N4148 3p	1N4149 4p	CP140 100 1N4002 1A/400V rectifier

Fibre-Optic Lasers

Two new CW-operated injection laser modules and a new series of long wavelength transimpedance preamplifier photodiode modules have been announced by RCA, designed specifically for fibre-optic communications systems.

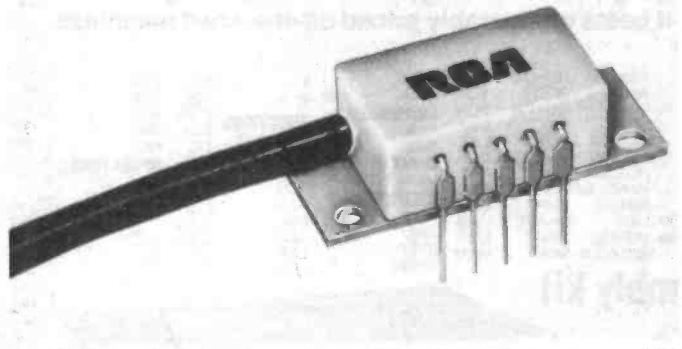
The C86041E (pictured below) is a gallium-aluminium-arsenide CW laser module with an output wavelength of 820 nanometers which is well-matched with the wavelength of silicon photodiodes. The C86042E is an indium-gallium-arsenide-phosphide CW laser module with an output wavelength of 1300 nanometers which is well-

matched with the wavelength of InGaAs/InP photodiodes. Both types are constructed with a 0.5 meter length of fibre-optic cable internally coupled to the emitting region of the laser chip and are supplied in a special dual-in-line package for simple mounting and good thermal dissipation. The package also contains a photodiode which monitors the laser output, and which may be used in a negative feedback circuit to stabilise the output of the laser. An RTD temperature detector is provided for high duty-cycle pulsed operation at case temperatures up to 70°C.

The C30986E Series are high speed InGaAsP (p-i-n) photodiodes with a hybrid preamplifier supplied in a 14-pin dual-in-line package. A glass win-

dow provides optical access to the photodiode and a 50 um graded index fibre pigtail is included. This device provides high responsivity between 900 and 1700 nanometers and is optimised for the detection of 1300 and 1550 nm sources.

Additional information may be obtained from RCA Electro-Optics and Devices, Lincoln Way, Windmill Road, Sunbury-on-Thames, Middx. TW16 7HW, tel 09327 85511.



Motor Speed Controller

Ferranti Electronics has recently introduced the ZN411E which, with a minimum of external components, provides precise speed control for electric motors.

Originally designed for use in a professional power drill made by a well known UK power tool manufacturer, the ZN411 will operate from the AC mains or a suitable DC supply and has an on-chip shunt regulator. The circuit has a power down reset

facility and a 'soft start' capability whereby the speed builds up smoothly to the set speed. It produces negative triac firing pulses and has a triac retrigger facility. A reversing input on the chip will stop the motor, which then goes through the soft start to reach the speed set for the reverse direction.

The device is available in an 18 lead plastic Dual in Line package. Details from Ferranti Electronics Limited, Fields New Road, Chadderton, Oldham, Lancashire, OL9 8NP, tel 061-624 0515.

New Op-Amps

Burr-Brown International have introduced two new operational amplifiers — a high power hybrid and a low bias current monolithic device with a JFET input.

The OPA501 is rated for continuous 80 W operation and will withstand peaks of at least 200 W without damage. It can deliver 10 amps peak into a 2 ohm load, 4 amps continuously into 5 ohms, and has output current limiting to protect both the amplifier and the load in the event of excessive drive or a fault occurring. Unity gain bandwidth is 1 MHz and the full power bandwidth is typically 16 kHz. Common mode input impedance is 250 M ohms and differential configuration input impedance is 10 M ohm.

The device is housed in an eight pin TO-3 package, the outer can being electrically isolated in order to simplify heat sinking (why aren't more power devices?). The OPA501 is designed to operate on supplies from ± 10 V to ± 36 V, and is very tolerant of supply voltage variations. Suggested applications include servo amplifiers, actuator controls and audio amplifiers.

The OPA100 JFET monolithic operational amplifier has a bias current of less than 0.5 pA at room temperature and less than 1 pA at 75°C but does not sacrifice performance in other areas to achieve this. Maximum offset voltage is 250 uV, offset voltage drift is 5 uV/°C maximum, and input impedance is 10^{12} ohms. All specifications are guaranteed over the temperature range -25°C to $+85^{\circ}\text{C}$.

Supplied in an eight-pin TO-99 package, the OPA100 uses the standard 741 type pin connections. Suggested applications include current-to-voltage converters, precision sample and hold amplifiers, and voltage amplification in high impedance circuits using biological probes, pH electrodes, etc. Burr-Brown International Ltd, Cassiobury House, 11-19 Station Road, Watford, Herts WD1 1EA.

Small Memory

WANG Laboratories Inc, has introduced a low-cost high-density dynamic random access memory module. Called SIMM (single in-line memory module) the module more than quadruples the density of the memory that can be positioned within a specified printed circuit board area, using industry-standard mounting practices.

Measuring only .75 x 3.00 inches, the WANG SIMM integrates nine separate 64K DRAMs and related decoupling chip capacitors into 64K bytes of memory with parity (error detection). Suitable for either direct mounting or socketing on a PCB, the module offers the potential for clustering as much as one full megabyte of memory within a three-by-four-inch area of the PCB. WANG says that the SIMM could significantly lessen the need for 256K technology, since it is denser than 256K and is available now. WANG UK, 661 London Road, Isleworth, Middlesex TW7 4EH, tel 01-560 4151.

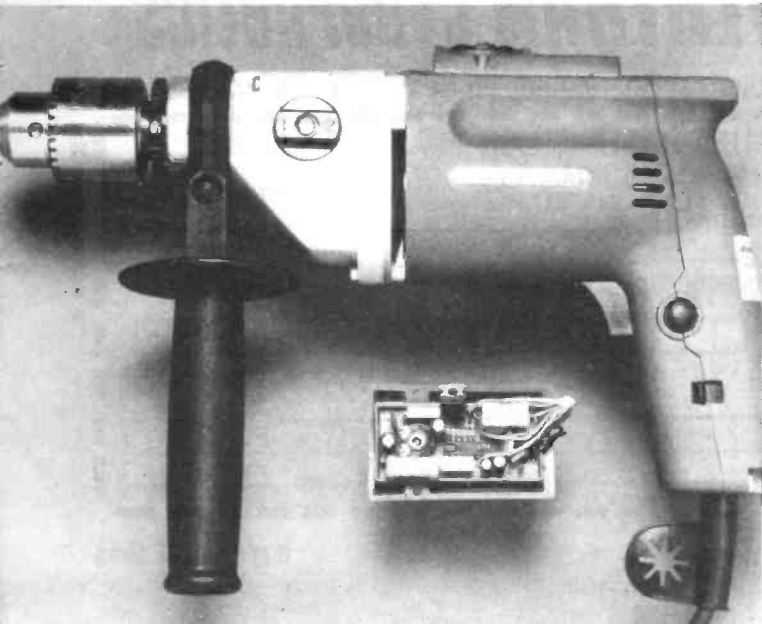
New CCDs From Plessey

Plessey have introduced four new CCD arrays: the MS1002-1 and 2 850 bit registers, and the MS1003-1 and 3 910 bit registers. All these devices are intended for video line storage, an application to which CCDs seem to be well suited.

The MS1002-1 is for analogue storage with interrupted clocks; the MS1002-3 is for delay-line operation with a continuous clock. Both have a video bandwidth of over 5 MHz with a clock frequency of 13.3 MHz, three times the PAL sub-carrier frequency.

The MS1003-1 and 3 are similar, but are intended for use with a 14.3 MHz clock, four times the NTSC colour sub-carrier frequency.

The advent of satellite TV, if it ever does happen, will increase the demand for storage facilities within receivers; as regular readers may recall, the C-MAC system chosen for UK transmissions involves a large degree of time-multiplexing of the TV signal, as opposed to frequency multiplexing which is the predominant form used at the moment. Plessey Semiconductors, Cheney Manor, Swindon, Wilts SM2 1QW, tel 0793 36251.





BUILD YOUR OWN CORTEX 16 bit, 64 RAM colour computer

Standard features –

- High speed 24K byte extended basic interpreter
- Powerful TMS9995 16 bit microprocessor
- 48 bit floating point gives 11 digit accuracy
- High resolution (256 x 192) colour graphics
- Screen memory does not use up user memory space
- 16 colours available on the screen together in graphic mode
- Fast line drawing and point plotting basic commands
- High speed colour shape manipulation from basic
- Full textual error messages
- String and Array size limited only by memory size
- Real time clock included in basic
- Interval timing with 10mS resolution via TIC function
- Named load and save of basic or machine code programs
- Auto-run available for any program
- Powerful machine code monitor
- Assembler and Disassembler included as standard
- Auto line numbering facility
- Full renumber command
- Simple but powerful line editor
- Buffered i/o allows you to continue executing the program while still printing
- Flexible CALL statement allows linkage to machine code routines with up to 12 parameters
- Basic programs may contain spaces between key words to make programs readable without using more memory
- Over 34K bytes available for basic programs
- Extended basic includes IF-THEN-ELSE
- Supports up to 16 output devices: Screen and cassette interfaces included as standard
- Supports bit manipulation of variables from basic
- Error trapping to a basic routine included
- Basic supports Hexadecimal numbers
- Separate 16K video RAM for graphics

With this powerful machine (featured in Electronics Today International as a constructional project) you have access to highly advanced systems and software developed specially by MPE Ltd for the CORTEX. For business, education, R & D – or simply increasing your knowledge and understanding of computers – it beats comparably priced off-the-shelf machines hands down!

STATEMENTS	PRINT	TIME	RENUM	MAG	MWD	:	(INT	POS
IF	?	WAIT	BOOT	TOF	BASE	@)	LOG	COL
ELSE		SAVE	GRAPH	TON	COMMANDS	#	FUNCTIONS	SOR	MOD
ON		LOAD	TEXT	DIM	RUN		FNA-FNZ	SYS	RND
GOTO	1 UNIT	MOTOR	PLOT	LET	SIZE	?	ASC	TIC	KEY
GOSUB	BAUD	ESCAPE	UNPLOT	DEF	CONT	?	ADR	SGN	OPERATORS
POP	CALL	NOESC	COLOUR	NEW	MON	\$	SIN	BIT	OR
REM	DATA	RANDOM	CHAR	END		%	ATN	CRB	LOR
FOR	READ	ENTER	SPRITE	BIT	DELIMITERS	.	SIN	CRF	AND
NEXT	RESTOR	LIST	SHAPE	CRB	TO	:	EXP	MEM	LAND
ERROR	RETURN	PURGE	SPUT	CRF	TAB	!	COS	MWD	NOT
INPUT	STOP	NUMBER	SGET	MEM	STEP	&	EXP	LEN	LNOT
					THEN		FRA	MCH	LXOR

Self assembly kit

£295

Ready built £395

All prices exclusive of VAT. Carriage paid.



Optional extras

RS232C interface kit	£9.20	Ready built	
Floppy disc interface	£65.50	CORTEX B – Basic machine	
Pair of 5 1/4" disc drives	£300.00	+ RS232C	£410.00
Hardware kit & connectors	£49.50	CORTEX C – as above + disc drives	£895.00

Full assembly instructions and 216 page user's manual.

POWERTRAN cybernetics

Portway Industrial Estate, Andover SP10 3ET. Tel: 0264 64455

To POWERTRAN CYBERNETICS, Portway Industrial Estate, Andover, Hants SP10 3NM.

Please send me _____

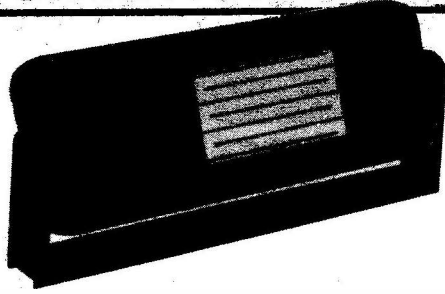
I enclose cheque for _____ or charge to:

Access/Barclaycard A/C No.

Name _____

Address _____

Tel. _____



Competition

Regisbrook of Reading have challenged the electronics world to a test of imagination with an astonishing new device — the problem is that they can't think of any use for it!

The device is a highly effective portable humidity sensor — so sensitive that it will react to the moisture in exhaled breath at a distance of two feet. It is battery powered and pocket sized, with an adjustable mounting bracket and no on/off switch. The unit features a sensing grid of interleaved gold filaments on a ceramic substrate, a robust plastic case, and hermetically sealed electronics for long term stability. It has an integral alarm

bleeper which cannot be triggered by short circuiting the gold grid and which stops when the sensing grid is wiped dry.

A grand prize of a magnum of champagne, and ten second prizes of the actual moisture sensors are offered by Regisbrook for the best suggestions (serious or imaginative) of what to do with the things. Ever anxious to be of service (and to assist in the disposal of alcoholic beverages) ETI would like to suggest that a few samples be sent to Conservative Central Office for use in the selection of suitably 'dry' parliamentary candidates. Readers with better suggestions should contact Regisbrook Ltd., Studio House, 215 Kings Road, Reading, Berks.

Improve Your Knowledge Here

Calling all would-be ham! Bradford & Ilkley Community College are running a 1 year course in preparation for the Radio Amateurs Examination. Enrolment starts on 6th September. Those interested should contact P. Nurse, Course Tutor, Bradford & Ilkley Community College, Great Horton Road, Bradford, West Yorkshire BD7 1AY.

For those unable to get to Bradford, Frank Fear is running a similar course at Barr Beacon Comprehensive School, near the M5/M6 junction. Enrolment is on 22nd September and there's even a special 10 week crash course for those with basic electrical knowledge. Further details, telephone Aldridge 52706.

If amateur radio isn't your scene why not try Information

Technology? Thames Polytechnic's new four year honours degree course starts in October and includes a year's industrial experience. Details from the Academic Registrar, Thames Polytechnic, Wellington Street, Woolwich, London SE18 6PF.

Salford University are running a series of one and two day courses on various aspects of the CBM/PET microcomputer. The first course starts on 13th September and the complete series will be repeated in three months time. For details, telephone 061-736 5843, extension 248.

Would-be movie directors should take their megaphones and folding chairs down to Piccadilly where JVC are holding two and three day courses in video production. The courses, which cater for both beginners and the more advanced, are repeated twice monthly. Dates, etc from Phil Compton or Mike Whyman at the JVC Video Information Centre, 82 Piccadilly, London W1, or telephone 01-491 3775.

Shorts

- Not one but two catalogues from Ambit! One is aimed at industrial and commercial users and comes free on request; the other is for lesser mortals and costs 80p from your friendly local newsagent. Ambit International, 200 North Service Road, Brentwood, Essex CM14 4SG.

- Concerned Technology, an exhibition of microprocessor based aids for the disabled and those with special needs, begins its twenty-nine stop tour in Hastings on September 8th. Details from Nancy Shawcross or Sue Hardwick on 01-789 4055/6.

- Silicon Valley, eat your heart out! The London borough of Islington is so proud of its local micro-electronics industries that it has made a film about them. "Silicon Green" runs for thirty minutes and is available to schools, business promotions, etc. Contact Jane Smith, 01-226 1234, ext. 268.

- NAHBO, the National Association of Hospital Broadcasting Organisations, is holding its autumn conference at the Ladbroke Mercury Hotel, Watford, on 14th/16th October. Exhibitors and individuals interested in Hospital Radio are welcome. NAHBO, 56 Fleet Road, Benfleet, Essex SS7 5JN.

- Sinclair Research are offering a ZX81, a 16K RAM pack, and a software cassette all for £45. Available for a two month trial period, the 'starter pack' will save you £30 on current prices and is on sale at Boots, John Menzies, and other Sinclair stockists. How long will it be before they start giving away computers with packets of breakfast cereal?

- The Intron IFG 422 is a 0.1 Hz to 2 MHz function generator offering sine, square, triangle, ramp, and pulse waveforms. It has 50 ohm and TTL outputs, is light and compact, and costs £195 from House of Instruments, Clifton Chambers, 62 High Street, Saffron Walden, Essex CB10 1EE, tel 0799 24922.

- Cable Check FC is a bench top go/no-go tester for insulation displacement type ribbon cables. It can test all cable/connector combinations up to 37 ways for both continuity and shorts. Cable Check Systems, Sanderson Centre, Lees Lane, Gosport, Hants, tel 07017 28396.

- The South Coast Hi-Fi Show will be taking its bucket and spade and an awful lot of audio goodies down to Brighton on the 11th, 12th and 13th November. Venue is the Royal Albion Hotel, Old Steine, and details are available from Tim Purcell, 137 Marina, St. Leonards-on-Sea, East Sussex TN38 0BT, tel 0424 715133.

- Market Logic's new Macrobyte printer buffer stores computer data output until the printer is ready for it, allowing the computer to carry on at its own speed. Designed to work with a Centronics interface, the Macrobyte comes in 16 k, 32 k and 64 k versions and prices start at £99.95. Tel 0432-70 456.

- Alcon Instruments claim their new pocket sized direct reading capacitance meter is as good as most bench-mounted bridges. The CP570 has five ranges from 0-50 pF up to 0-0.5 uF, a three inch meter scale, a basic accuracy of $\pm 3\%$ and costs £51.06 complete with leads and case. Alcon Instruments Ltd, 19 Mulberry Walk, London SW3, tel 01-352 1897.

- Advanced Micro Devices now have available a 256 K EPROM which has an access time of only 170 nS. The Am27256 is organised as 32 K \times 8 bits, requires a single 5 V supply, has TTL compatible inputs and outputs and uses the standard 28 pin configuration. Advanced Micro Devices (UK) Ltd, AMD House, Goldsworth Road, Woking, Surrey GU21 1JT, tel 04862 22121.

- For those whose penchant is listening to Stockhausen in the sauna, Fuji have introduced a high temperature cassette tape. The GT-1 will withstand 110°C (230°F) without damage to either shell or tape and is intended for use in cars. Bell & Howell, Alper-

ton House, Bridgewater Road, Wembley, Middlesex HA0 1EG, 01-902 8812.

- Just when you'd got all your accessories colour coordinated, Hanimego and introduce their LC751. Described as a high fashion calculator, it is superslim and has a glossy black lacquer finish highlighted by gold keys and trim. And when fashions change, you can always use it to work things out on!

- With one snap of their powerful jaws, 3M's new TH212 and TH213 will strip both solid and stranded wires without damage to the conductor. The TH212 covers wires from 0.2 mm to 0.9 mm and the TH213 covers wires from 0.75 to 6 mm, both types adjusting automatically to all wire sizes within their range. 3M UK, 3M House, P.O. Box 1, Bracknell, Berkshire RG12 1JU, tel 0344 26726.

- Roxburgh Suppressors Ltd have introduced two mains filters designed for direct mounting onto printed circuit boards. The PC103 and PC105 are rated at 3 and 5 amps respectively and are full encapsulated. Roxburgh Suppressors Ltd, Eagle Road, Rye, East Sussex, tel 07973 3725.

- An Apple a day has clearly done wonders for the California based computer company. They've just produced the one millionth Apple personal computer and to celebrate are giving an Apple IIe to every school in the state.

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AUTO-ELECTRONIC PRODUCTS

KITS OR READY BUILT

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Performance at only 6 volts (max. supply 16 volts)
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TYPEWRITER INTERFACE

Daft, isn't it? You can't get a daisy-wheel printer for less than about £400, but you can buy daisy-wheel typewriters containing not just the printer, but a keyboard and some electronics for just over £200. Jon Tyler shows us how you can interface one to your micro.

The Silver Reed EX 42 typewriter is logically divided into two parts. The keyboard is controlled by its own 8049 microcontroller while the printing mechanism is controlled by an 8039 with an EPROM containing the software appropriate to printing the required character set. Information is passed between these devices using a serial interface. It is possible to interface to the machine via an edge connector on the main PCB but if it is to be used as an output device only, a simpler interface may be constructed which operates in parallel with the existing keyboard.

The keyboard is arranged as a 8 × 8 array of keyswitches which are positioned according to the normal QWERTY convention. There are 25 connections made between the keyboard and the electronics, these taking the form of one 10-way and one 15-way ribbon cable connector. The modification to the typewriter is to disconnect these and reconnect them through suitable plug-socket pairs. In the prototype, these connectors were made from a 40-way wire-wrap IC socket cut into suitable lengths.

Construction

The prototype was constructed in a plastic box measuring about 3" by 4.5" by 1.5". A 26-way ribbon cable and matching connector was used to connect to the typewriter and a 12-way connector used to connect to the micro-computer parallel port. The ribbon cable was connected to the typewriter using the home-made adaptors shown in Fig. 5. These were made by cutting a 40-way wire wrap socket into 10

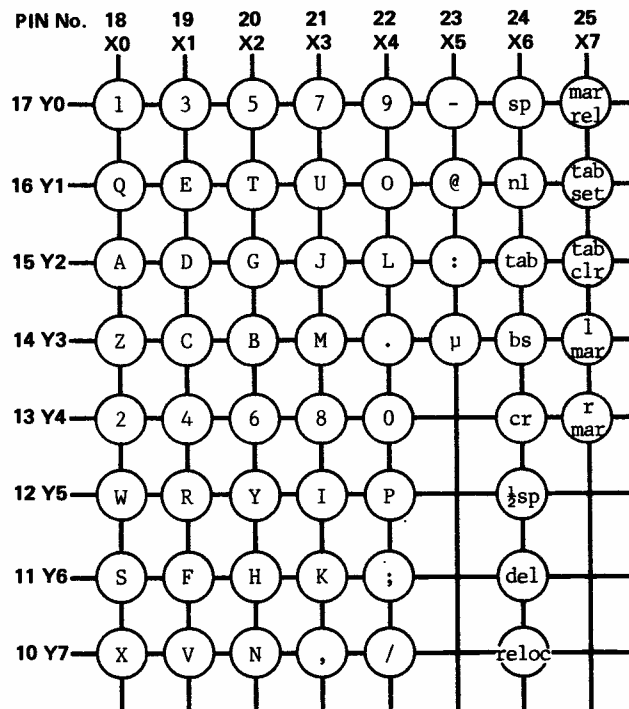
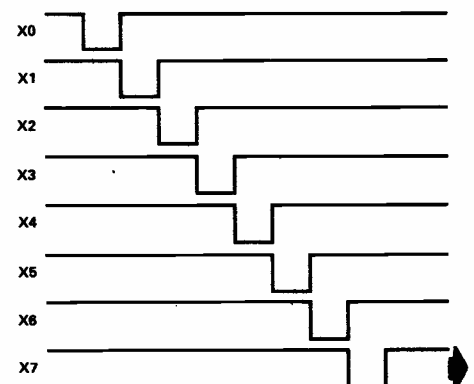


Fig. 1 (above) Keyboard matrix with pin connections.

Fig. 2 (right) How the typewriter scans the input multiplex lines to the keyboard.

and 15 way lengths which were soldered to strips of Veroboard. The ribbon cable was then soldered to these adaptors, the 26th way being left free and fitted with a solder tag.

The typewriter cover is removed by unscrewing the two grub screws at either end of the platen (roller) using an Allen key. Remove the



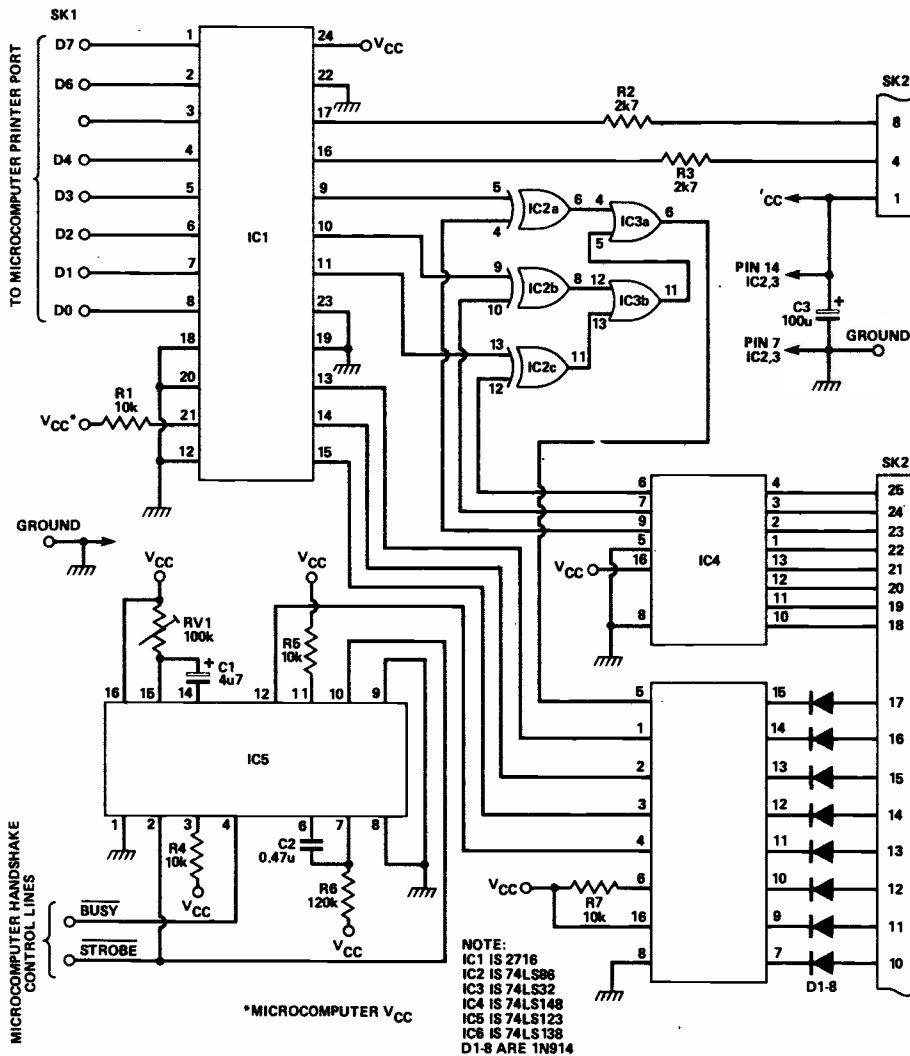
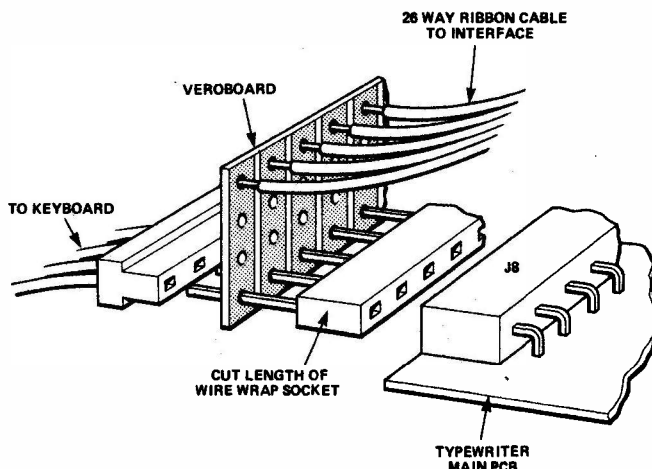


Fig. 3 Circuit diagram for the interface.

BUYLINES

The only difficult item here is the wire-wrap socket. We suggest you experiment with any odd DIL wire-wraps you may have around before splashing out on anything fancy. 26 way ribbon cable is easily obtained by chopping up 30 or 34 way, etc. RV1 is available from Maplin,

the 26 way PCB plug and socket from Ambient, the 2716 and other semiconductors are available from several of our advertisers, and the PCB can be obtained from you-know-where (but see page 77).



HOW IT WORKS

The microcomputer is programmed to put out an 8 bit ASCII coded character via a parallel port. The codes used to denote the X and Y position of the typewriter keys do not correspond to the ASCII codes, and upper and lower case character are selected by pressing the 'shift' key. Each character to be printed by the microcomputer must therefore be translated into the required 3 bit X and 3 bit Y code together with the one bit shift code. This may be achieved either in the microcomputer software (see Fig. 6) or alternatively by programming a suitable EPROM. A listing of the required codes is given in Table 1.

The eight 'X' lines X0 to X7 are scanned by the typewriter as shown in Fig. 2. These lines are connected to IC1 in the interface which encodes the 8 line signal into a 3 bit code. The three outputs A0, A1 and A2 from this priority encoder form the logical inverse of the 3 bit column number which corresponds to the 'X' line currently enabled. These three bits are compared with the least significant 3 bits of the output byte using IC3 and IC4 which perform an exclusive-OR operation. When a match occurs, pin 6 of IC4 goes low and this is used to control the 'Y' outputs.

The pressing of a key is detected in the typewriter by reading in the eight-bit value on the 'Y' lines and comparing this with the previous value. When a key is pressed, the corresponding 'Y' line, which is normally held at '1' by a pull-up resistor, will be made '0' when the appropriate 'X' line is enabled. The typewriter saves the current 'X' (column) code together with the 'Y' (row) code whenever a 'Y' line goes low. To simulate this, a 3 line to 8 line decoder, IC2, is used. The select inputs A, B and C are connected to bits 3, 4 and 5 of the output from the microcomputer parallel port.

The shift operation is invoked by pressing the shift key, which is a normally open switch separate from the scanned keys. This switch operates the interrupt line INT to the 8049 microcontroller in the typewriter and is normally held low. In the interface it is connected to bit 6 of the parallel output. There is a similar arrangement to select the 'alternate character set' which is required for 3 of the ASCII codes. This function was assigned to bit 7.

The typewriter manufacturers give a maximum printing speed of nine characters per second. A monostable, IC5, is used to achieve the correct operating speed. IC5 is triggered by an output strobe line which in the prototype was arranged to be active on the positive going edge.

After a valid character has been output from the microcomputer, the strobe line triggers the monostables, and the Q₁ output is used with the output of the exclusive-OR circuit to activate the 3 line to 8 line decoded IC2. The monostable output Q₂ is used as a busy signal by the microcomputer.

Fig. 4 (left) This shows how the improvised connector is made and how it is fitted into the typewriter.

PARTS LIST

RESISTORS (5% 1/4W)
 R1,4,5,7 10k
 R2,3 2k7
 R6 120k
 RV1 100k multi-turn

CAPACITORS (10V working min)
 C1 4µ7 tant (see text)
 C2 470n polyester
 C3 100µ tant

SEMICONDUCTORS
 IC1 2716 EPROM
 IC2 74LS86
 IC3 74LS32
 IC4 74LS148
 IC5 74LS123
 IC6 74LS138
 D1-8 1N914 or similar

MISCELLANEOUS
 26-way PCB mounting plug and socket (BICC-vero 901-71105D or similar);
 40-way wire-wrap socket (see text); PCB;
 ribbon cable; connector and cable to suit microcomputer used.

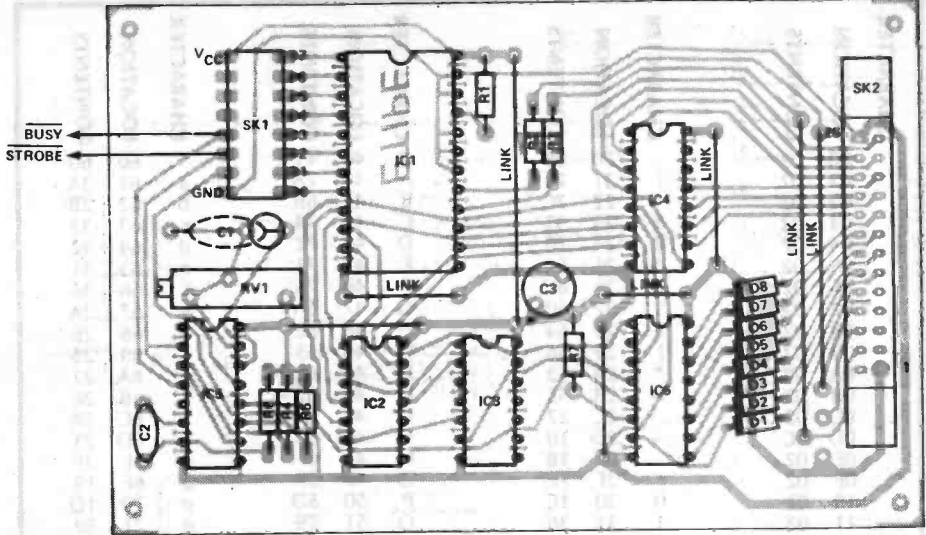


Fig. 5 (above) Overlay diagram of the interface PCB.

Fig. 6 (below) Program to drive the interface, to be used when an EPROM is not used.

```

10 REM PRINTER SIMULATOR ROUTINE
20 DIM D(128)
30 FOR N=1 TO 128
40 READ D(N)
50 NEXT N
60 INPUT "FILENAME ",BE : REM READ A FILENAME
70 OPEN 1 BE : ON EOF GOTO 340
80 GET 1 CE
90 FOR P=1 TO LEN(CE) : REM GET EACH CHARACTER
100 AE=MID$(CE,P,1)
110 GOSUB 200
120 IF AE=CHR$(13) GOTO 140 : REM CHECK FOR CR
130 GOTO 180
140 AE=CHR$(10) : REM IF SO PUT IN A LF
150 GOSUB 200
160 FOR N=1 TO 125 : REM AND A DELAY
170 NEXT N
180 NEXT P
190 GOTO 80
200 X=D(ASC(AE)+1) : REM SUBROUTINE TO PRINT A CHARACTER
210 OUT(16R7E)=X : REM OUTPUT TO A SUITABLE PORT
220 OUT(16R7F)=0 : REM ISSUE THE STROBE
230 OUT(16R7F)=1
240 Y=IN(16R7F) : REM CHECK FOR BUSY
250 Y=MOD(Y,2)
260 IF Y<>0 GOTO 240
270 RETURN
280 DATA 2,2,2,2,2,2,2,2,2,2,2,9,2,2,12,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2,2
290 DATA 8,120,124,83,112,104,108,96,100,88,82,94,39,16,27,31,28,56,60,48
300 DATA 52,40,44,32,36,24,18,30,231,92,220,95,17,122,107,115,114,113,118
310 DATA 106,110,101,98,102,90,99,111,89,93,121,117,126,105,97,119,125
320 DATA 127,109,123,209,8,145,8,80,96,58,43,51,50,49,54,42,46,37,34,38,26
330 DATA 35,47,25,29,57,53,62,41,33,55,61,63,45,59,8,159,8,8,2
340 CLOSE 1
350 STOP
360 END
    
```

paper pan which is clipped in position under the platen, then unscrew the four screws in the base and remove the top cover by gently raising it at the back and pulling it forward. The keyboard may then be removed from its clips and turned upside down so as to gain access to the ribbon cable connectors. These are then removed from their receptacles on the PCB and the adaptors fitted. The 26th way from the ribbon cable must be taken to ground; a suitable point exists on the typewriter PCB marked 'GND'. The typewriter should then be reassembled and the interface connected to the microcomputer.

The period of the monostable must now be adjusted. Set up the microcomputer to send a reasonably long text file and adjust the preset (RV1) to give a printing rate of about nine characters per second. Allowance must be made for the time taken to perform a carriage return and line feed. If this is not done in the microcomputer's standard software (normally by sending an appropriate number of 'null' characters) then CR and/or LF must be detected in the interface program and suitable delays introduced. (See BASIC program, Fig. 4.)

Note that if you decide to use your micro to do the character coding rather than the EPROM (see How It Works and Fig. 6), then the EPROM should be omitted and what would have been its pins 1 to 8 should be linked to pins 17, 16, 15, 14, 13, 11, 10 and 9 respectively. If you want to retain the option of using an EPROM at a later

PROJECT : Typewriter Interface

CHARACTER	LOCATION	CONTENTS	CHARACTER	LOCATION	CONTENTS	CHARACTER	LOCATION	CONTENTS
00	02		20	08		@	40	11
01	02	!	21	78		A	41	7A
02	02	"	22	7C		B	42	6B
03	02	\$	23	53		C	43	73
04	02	%	24	70		D	44	72
05	02	&	25	68		E	45	71
06	02	'	26	6C		F	46	76
07	02	(27	60		G	47	6A
08	02)	28	64		H	48	6E
09	02	*	29	58		I	49	65
0A	09	+	2A	52		J	4A	62
0B	02	,	2B	5E		K	4B	66
0C	02	;	2C	27		L	4C	5A
0D	0C	:	2D	10		M	4D	63
0E	02	?	2E	1B		N	4E	6F
0F	02	/	2F	1F		O	4F	59
10	02	0	30	1C		P	50	5D
11	02	1	31	38		Q	51	79
12	02	2	32	3C		R	52	75
13	02	3	33	30		S	53	7E
14	02	4	34	34		T	54	69
15	02	5	35	28		U	55	61
16	02	6	36	2C		V	56	77
17	02	7	37	20		W	57	7D
18	02	8	38	24		X	58	7F
19	02	9	39	18		Y	59	6D
1A	02	:	3A	12		Z	5A	7B
1B	02	;	3B	1E		[5B	D1
1C	02	?	3C	e)]	5C	08
1D	02	-	3D	5C		}	5D	91
1E	02	=	3E	DB		—	5E	08
1F	02	>	3F	3F			5F	50

Table 1 EPROM contents and what they do.

stage, then perhaps the best way of doing this is with a header plug in the IC socket.

Once source of trouble that you may have is with the stability of C1. We've specified a tantalum type, but you may find it necessary to use a polystyrene (we know, we know, they're big and expensive). The exact value of C1 isn't important, but it must not vary too much over time or with temperature and electrolytics are capable of doing just that.

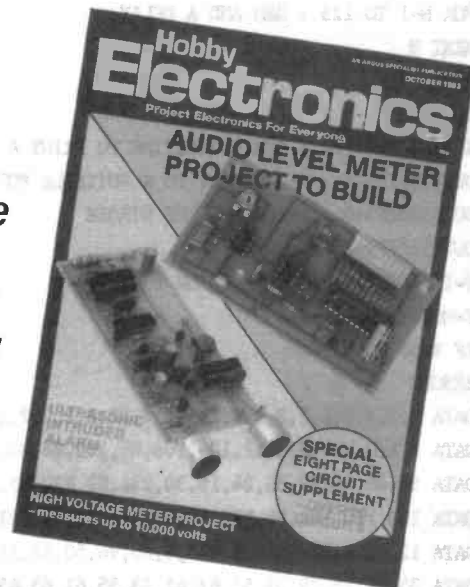
One option you might like to use is to connect two on off toggle switches between the 5 V rail and pins 6 and 8 of the connector SK2. These allow the selection of 15 characters per inch type spacing and of an alternative keyboard to be selected, both independent of the computer output.

This isn't necessarily the end of the story. We're so impressed by the keyboard on the EX42 that we're looking at ways of using it too. Watch this space!

ETI

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12.0 1.2V
15.0 1.5V

2.5 x 17 2.90
3.75 x 17 3.85
4.75 x 17 4.93
10 Board 1.92
15 way 1.03
25 way 1.60
37 way 2.50

CONNECTIONS
R3 9x5.2x 3.16
R4 11x6x3 3.16
R5 11x7x4 3.85
R6 13x8x4 4.76
R7 15x8x4 7.35

VEROBXES
Black Plastic
V2 120x80x
15mm 90p
V3 180x110x55
1.60

EURO CONNECTORS
31 way 1.75
64w A-B
Straight 2.25
64w A-B
Angled 2.40

SLOPING FRONT
VEROBXES
V2 220x174
15mm 100.40
V3 171x121x
75mm 5.65

REMOVAL
CONTROL
(Handheld) BOX
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White 89p

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Grey (Suffix G)

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16 way 1.25
20 way 1.49
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Dual 1.20
Quad 2.50

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CONNECTORS
PP3 Snap 25p
PP3 Snap 12p
Box for 4HP1 27p

RESISTORS

IDEAL AS BEAMS

100 3V 32p

2.5 x 17 2.90

CONNECTIONS

VEROBXES

EURO CONNECTORS

SLOPING FRONT

REMOVAL

PROFESSIONAL

IDC CONNECTORS

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PHONO PLUGS

PHONO CHAS

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BF201 24p	ME9005 60p	TIS48 50p
BF202 27p	ME9006 65p	TIS49 50p
BF203 30p	ME9007 70p	TIS50 50p
BF204 33p	ME9008 75p	TIS51 50p
BF205 36p	ME9009 80p	TIS52 50p
BF206 39p	ME9010 85p	TIS53 50p
BF207 42p	ME9011 90p	TIS54 50p
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BF209 48p	ME9013 1.00	TIS56 50p
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BF216 69p	ME9020 1.35	TIS63 50p
BF217 72p	ME9021 1.40	TIS64 50p
BF218 75p	ME9022 1.45	TIS65 50p
BF219 78p	ME9023 1.50	TIS66 50p
BF220 81p	ME9024 1.55	TIS67 50p
BF221 84p	ME9025 1.60	TIS68 50p
BF222 87p	ME9026 1.65	TIS69 50p
BF223 90p	ME9027 1.70	TIS70 50p
BF224 93p	ME9028 1.75	TIS71 50p
BF225 96p	ME9029 1.80	TIS72 50p
BF226 99p	ME9030 1.85	TIS73 50p
BF227 1.02	ME9031 1.90	TIS74 50p
BF228 1.05	ME9032 1.95	TIS75 50p
BF229 1.08	ME9033 2.00	TIS76 50p
BF230 1.11	ME9034 2.05	TIS77 50p
BF231 1.14	ME9035 2.10	TIS78 50p
BF232 1.17	ME9036 2.15	TIS79 50p
BF233 1.20	ME9037 2.20	TIS80 50p
BF234 1.23	ME9038 2.25	TIS81 50p
BF235 1.26	ME9039 2.30	TIS82 50p
BF236 1.29	ME9040 2.35	TIS83 50p
BF237 1.32	ME9041 2.40	TIS84 50p
BF238 1.35	ME9042 2.45	TIS85 50p
BF239 1.38	ME9043 2.50	TIS86 50p
BF240 1.41	ME9044 2.55	TIS87 50p
BF241 1.44	ME9045 2.60	TIS88 50p
BF242 1.47	ME9046 2.65	TIS89 50p
BF243 1.50	ME9047 2.70	TIS90 50p
BF244 1.53	ME9048 2.75	TIS91 50p
BF245 1.56	ME9049 2.80	TIS92 50p
BF246 1.59	ME9050 2.85	TIS93 50p
BF247 1.62	ME9051 2.90	TIS94 50p
BF248 1.65	ME9052 2.95	TIS95 50p
BF249 1.68	ME9053 3.00	TIS96 50p
BF250 1.71	ME9054 3.05	TIS97 50p
BF251 1.74	ME9055 3.10	TIS98 50p
BF252 1.77	ME9056 3.15	TIS99 50p
BF253 1.80	ME9057 3.20	TIS100 50p
BF254 1.83	ME9058 3.25	TIS101 50p
BF255 1.86	ME9059 3.30	TIS102 50p
BF256 1.89	ME9060 3.35	TIS103 50p
BF257 1.92	ME9061 3.40	TIS104 50p
BF258 1.95	ME9062 3.45	TIS105 50p
BF259 1.98	ME9063 3.50	TIS106 50p
BF260 2.01	ME9064 3.55	TIS107 50p
BF261 2.04	ME9065 3.60	TIS108 50p
BF262 2.07	ME9066 3.65	TIS109 50p
BF263 2.10	ME9067 3.70	TIS110 50p
BF264 2.13	ME9068 3.75	TIS111 50p
BF265 2.16	ME9069 3.80	TIS112 50p
BF266 2.19	ME9070 3.85	TIS113 50p
BF267 2.22	ME9071 3.90	TIS114 50p
BF268 2.25	ME9072 3.95	TIS115 50p
BF269 2.28	ME9073 4.00	TIS116 50p
BF270 2.31	ME9074 4.05	TIS117 50p
BF271 2.34	ME9075 4.10	TIS118 50p
BF272 2.37	ME9076 4.15	TIS119 50p
BF273 2.40	ME9077 4.20	TIS120 50p
BF274 2.43	ME9078 4.25	TIS121 50p
BF275 2.46	ME9079 4.30	TIS122 50p
BF276 2.49	ME9080 4.35	TIS123 50p
BF277 2.52	ME9081 4.40	TIS124 50p
BF278 2.55	ME9082 4.45	TIS125 50p
BF279 2.58	ME9083 4.50	TIS126 50p
BF280 2.61	ME9084 4.55	TIS127 50p
BF281 2.64	ME9085 4.60	TIS128 50p
BF282 2.67	ME9086 4.65	TIS129 50p
BF283 2.70	ME9087 4.70	TIS130 50p
BF284 2.73	ME9088 4.75	TIS131 50p
BF285 2.76	ME9089 4.80	TIS132 50p
BF286 2.79	ME9090 4.85	TIS133 50p
BF287 2.82	ME9091 4.90	TIS134 50p
BF288 2.85	ME9092 4.95	TIS135 50p
BF289 2.88	ME9093 5.00	TIS136 50p
BF290 2.91	ME9094 5.05	TIS137 50p
BF291 2.94	ME9095 5.10	TIS138 50p
BF292 2.97	ME9096 5.15	TIS139 50p
BF293 3.00	ME9097 5.20	TIS140 50p
BF294 3.03	ME9098 5.25	TIS141 50p
BF295 3.06	ME9099 5.30	TIS142 50p
BF296 3.09	ME9100 5.35	TIS143 50p
BF297 3.12	ME9101 5.40	TIS144 50p
BF298 3.15	ME9102 5.45	TIS145 50p
BF299 3.18	ME9103 5.50	TIS146 50p
BF300 3.21	ME9104 5.55	TIS147 50p

BRIDGE

Watt type	2 5 Watt
E24 Series	E24 Series
7 5 Watt	7 5 Watt 1 10

SW wired ended following cones only: 3V3, 3V6, 4V3, 4V7, 5V6, 7V5, 7V7, 8V1, 8V1, 10V, 12V, 20V, 33V, 51V, 62V, 68V, 1.25

2 5 Watt	CA3033 5.44
E24 Series	CA3036 5.18
7 5 Watt	CA3039 5.10
LM1305	LM1307 3.70
LM1310N	LM1310N 1.48
LM1308	LM1308 2.25
LM1458	LM1458 4.46
LM1468	LM1468 4.46
LM1800	LM1800 3.24
LM1801	LM1801 2.99
LM1812	LM1812 4.60
LM1818	LM1818 4.00
LM1820	LM1820 2.15
LM1845	LM1845 4.42
LM1846	LM1846 4.42
LM1850	LM1850 2.75
LM1851	LM1851 3.25
LM1852	LM1852 3.30
LM1853	LM1853 3.30
LM1854	LM1854 3.30
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LM1872	LM1872 3.30
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LM1893	LM1893 3.30
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LM1895	LM1895 3.30
LM1896	LM1896 3.30
LM1897	LM1897 3.30
LM1898	LM1898 3.30
LM1899	LM1899 3.30
LM1900	LM1900 3.30

OPTO

LED LAMPS	R = Red
	G = Green
	B = Blue
	Y = Yellow
	IR = Infra Red
CA3070	CA3070 1.35
CA3080	CA3080 1.35
CA3090	CA3090 1.35
CA3100	CA3100 1.35
CA3110	CA3110 1.35
CA3120	CA3120 1.35
CA3130	CA3130 1.35
CA3140	CA3140 1.35
CA3150	CA3150 1.35
CA3160	CA3160 1.35
CA3170	CA3170 1.35
CA3180	CA3180 1.35
CA3190	CA3190 1.35
CA3200	CA3200 1.35
CA3210	CA3210 1.35
CA3220	CA3220 1.35
CA3230	CA3230 1.35
CA3240	CA3240 1.35
CA3250	CA3250 1.35
CA3260	CA3260 1.35
CA3270	CA3270 1.35
CA3280	CA3280 1.35
CA3290	CA3290 1.35
CA3300	CA3300 1.35

DIODES

BR100	BR100 25p
BR110	BR110 25p
BR120	BR120 25p
BR130	BR130 25p
BR140	BR140 25p
BR150	BR150 25p
BR160	BR160 25p
BR170	BR170 25p
BR180	BR180 25p
BR190	BR190 25p
BR200	BR200 25p
BR210	BR210 25p
BR220	BR220 25p
BR230	BR230 25p
BR240	BR240 25p
BR250	BR250 25p
BR260	BR260 25p
BR270	BR270 25p
BR280	BR280 25p
BR290	BR290 25p
BR300	BR300 25p
BR310	BR310 25p
BR320	BR320 25p
BR330	BR330 25p
BR340	BR340 25p
BR350	BR350 25p
BR360	BR360 25p
BR370	BR370 25p
BR380	BR380 25p
BR390	BR390 25p
BR400	BR400 25p

THYRISTORS

CA3240E	CA3240E 1.04
CA3250E	CA3250E 1.04
CA3260E	CA3260E 1.04
CA3270E	CA3270E 1.04
CA3280E	CA3280E 1.04
CA3290E	CA3290E 1.04
CA3300E	CA3300E 1.04
CA3310E	CA3310E 1.04
CA3320E	CA3320E 1.04
CA3330E	CA3330E 1.04
CA3340E	CA3340E 1.04
CA3350E	CA3350E 1.04
CA3360E	CA3360E 1.04
CA3370E	CA3370E 1.04
CA3380E	CA3380E 1.04
CA3390E	CA3390E 1.04
CA3400E	CA3400E 1.04
CA3410E	CA3410E 1.04
CA3420E	CA3420E 1.04
CA3430E	CA3430E 1.04
CA3440E	CA3440E 1.04
CA3450E	CA3450E 1.04
CA3460E	CA3460E 1.04
CA3470E	CA3470E 1.04
CA3480E	CA3480E 1.04
CA3490E	CA3490E 1.04
CA3500E	CA3500E 1.04

RECTANGULAR

CA3033	CA3033 5.44
CA3036	CA3036 5.18
CA3039	CA3039 5.10
CA3041	CA3041 3.47
CA3042	CA3042 3.47
CA3043	CA3043 2.75
CA3046	CA3046 6.80
CA3047	CA3047 4.60
CA3048	CA3048 2.15
CA3049	CA3049 3.21
CA3050	CA3050 2.82
CA3051	CA3051 3.80
CA3052	CA3052 2.90
CA3053	CA3053 2.90
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CA3055	CA3055 2.90
CA3056	CA3056 2.90
CA3057	CA3057 2.90
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CA3061	CA3061 2.90
CA3062	CA3062 2.90
CA3063	CA3063 2.90
CA3064	CA3064 2.90
CA3065	CA3065 2.90
CA3066	CA3066 2.90
CA3067	CA3067 2.90
CA3068	CA3068 2.90
CA3069	CA3069 2.90
CA3070	CA3070 1.35
CA3071	CA3071 1.35
CA3072	CA3072 1.35
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CA3081	CA3081 1.35
CA3082	CA3082 1.35
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CA3084	CA3084 1.35
CA3085	CA3085 1.35
CA3086	CA3086 1.35
CA3087	CA3087 1.35
CA3088	CA3088 1.35
CA3089	CA3089 1.35
CA3090	CA3090 1.35
CA3091	CA3091 1.35
CA3092	CA3092 1.35
CA3093	CA3093 1.35
CA3094	CA3094 1.35
CA3095	CA3095 1.35

AUDIO DESIGN PART 2

As a general classification, one can divide linear ICs into two broad categories: purpose-built or dedicated ICs, aimed at the performance of one specific application; and general purpose ICs, such as the operational amplifier types, which are designed for use as versatile building blocks in a wide range of circuit configurations. The first of these categories contains a host of useful devices, whose numbers increase daily, that are capable of doing a very wide range of jobs, from providing well stabilised, ripple free, power supplies, to complete radio systems or audio amplifiers in a single package.

As a rule, hi-fi circuit designers and their customers tend rather to look askance at such purpose-built ICs in audio applications, since these devices are intended mainly for use in mass-produced consumer hardware, to simplify and reduce the cost of manufacture of competitively priced domestic electronics. A hi-fi specification is not usually either part of the IC designer's brief or of the customer's cost expectation. Nevertheless, some of these dedicated ICs perform extremely well, and have found their way into some of the most prestigious of audio systems. One must, therefore, try not to harbour preconceived opinions about their potential quality, but rather to judge these devices on their individual merits.

So far as the circuitry associated with these dedicated ICs is concerned, although it is great fun to explore their internal construction and to work out schemes for using them in applications their designers had never envisaged, this is rather an exotic field, and full of pitfalls for the unwary. So, in general, it is prudent to stick fairly closely to the circuit applications and component values recommended by the manufacturers in their application data sheets (which one should make sure one gets with the device) since it will probably be difficult to improve greatly upon these recommendations. With general purpose ICs, these constraints upon the method of possible use do not exist, and a very wide range of circuit uses can be envisaged.

Basic IC Amplifier Layouts

Taking the op-amp gain block as a starting point, this will be familiar as a simple, fairly wide bandwidth amplifier unit, having two inputs — one phase-inverting, one non-inverting, an output pin, and two further pins for connection to a +ve and -ve DC voltage supply. Some ICs of this type (in fact, most of the packages which contain just one gain block) also have two further pins which can be used, when connected to an external trimmer potentiometer, to adjust the DC output level of the amplifier when this is used as a DC amplifier stage. Such an op-amp IC will normally be designed to work over supply voltage ranges from $\pm 1.5\text{v}$ to $\pm 15\text{v}$, or indeed, in a suitable circuit layout, from a single DC supply within the range 3-30 volts.

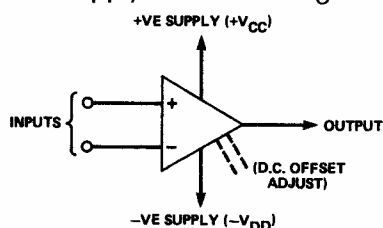


Fig. 1 Conventional circuit drawing of an operational amplifier.

The conventional circuit drawing for such an op-amp is shown in Fig. 1, where, as ever, the + and - symbols on the input leads imply the non-inverting and inverting amplifier characteristics. Although on this drawing I have shown the + and - supply connections to the IC, it is a common practice to take these as read where the ICs are used from a symmetrical or otherwise unremarkable power supply arrangement, and I propose to follow this convention and omit the power supply connections, where these are standard, in future drawings.

Most of the popular op-amps of the 741 type are what is known as *internally compensated*, which implies that negative feedback can be used in the circuit by the connection of a suitable network between the output and the inverting input pins, without having to worry about whether the amplifier will then be stable. In some of the earlier op-amp ICs, internal HF compensation was not provided, on the grounds that the necessary worst-case (unity-gain) compensation would lead to a less good HF performance from the IC, at higher than unity-gain conditions, than if the compensation was done by a suitably chosen network of Rs and Cs external to the IC package. However, this situation has been overtaken by progress in IC design, and most of the contemporary IC designs will give a good HF performance without the need to accept an inconvenient external RC network.

Typical values of *open-loop* (ie, before any negative feedback is applied) small signal, low frequency voltage gain are in the range 100,000-200,000, and the rejection of unwanted noise and voltage fluctuations on the voltage supply lines is usually of a similar order. The *common mode rejection ratio* (the ratio between the open loop gain and the 'fault' gain you get when you tie the + and - inputs together and use them as if they were just one input) is usually in this range also, but it is very difficult to organise any circuit layout which will allow gains much higher than 1000 to be achieved. The most common pin configurations for IC op-amps are shown in Fig. 2, and a typical circuit for a small signal AC voltage amplifier is shown in Fig. 3a.

The circuit will give a stage gain determined by the ratio $(R_3 + R_2)/R_2$. For the resistor values shown this is 48, and this circuit will have a frequency response at 1 V out

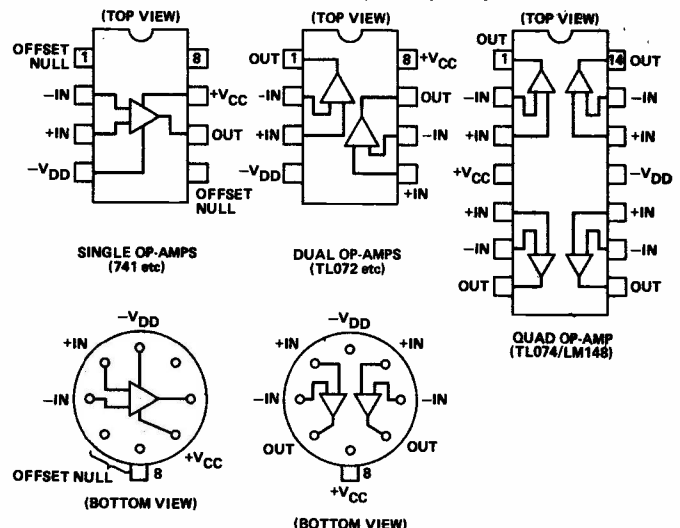


Fig. 2 Common op-amp pin connections.

Problems and ICs often, as we all know, go together. John Linsley Hood takes a look at the two of them in audio.

(RMS) — assuming a ± 15 V DC supply — which is substantially flat from 10 Hz to 30 kHz, the LF limit being set by the value chosen for C1 (bigger = lower) and the HF end being determined by the characteristics of the IC itself, as a consequence of its internal HF compensation.

At 1 V RMS and 1 kHz, a typical distortion figure into a 2k ohm load, would be about 0.02% and a S/N ratio (assuming a low source impedance) of about 75dB, when measured over the 10 Hz-30 kHz bandwidth. Substituting one of the newer designs of IC intended for audio use, such as a TL071 or a LF351, would halve the noise, increase the bandwidth to about 300 kHz, and reduce the 1 kHz distortion to some 0.002%. Such is progress!

Such a gain block, particularly when built using a TL071 or a LF351, both of which have high impedance FET inputs and very low noise characteristics, makes a very respectable hi-fi amplifier stage, in any application where a flat gain/frequency response is appropriate. To take advantage of the convenience of a single power supply rail, which facilitates joining IC circuitry on to discrete transistor layouts, the circuit of 3a can be rearranged as shown in Fig. 3b without any loss of performance. (To assist in comparing the layout of 3b with that of 3a, I have retained some of the component numbering of 3a in 3b.) Lower supply voltages diminish the available output swing and tend to worsen the THD (total harmonic distortion) at any given output, though this will only become conspicuous as the expected output signal level begins to get near the maximum available due to the DC supply provided.

Both of these op-amp gain stages compare favourably, both in terms of the cost of the components and in terms of performance, with the comparable separate transistor versions, the only major snag being the limit on the possible output voltage swing imposed by the ICs restricted HT supply capability. It is no use, therefore, to try to use an IC of this type if one wants a 100 V P-P signal output.

As shown in Fig. 2, IC op-amps of this type are available in packages which contain up to four separate amplifiers on the same chip, usually with very little sacrifice in performance, and with only minimal signal breakthrough from one to another. In particular, the dual op-amp TL072 and LF353 types have become very popular among audio circuit designers as a means of handling a pair of stereo signal channels in one device.

Frequency Response Shaping

While quite a lot of audio signal handling can be done with stages with linear gain/frequency characteristics, it is very useful to be able to modify this frequency response. I have mentioned above, in the case of Fig. 3a, that the LF response was determined by the value of C1. This is because, in a feedback amplifier, the gain of the stage is really determined by the ratio of the impedances in the two limbs of the feedback network. So long as the impedances of any capacitors (or inductors) in these limbs can be ignored in comparison with the resistive elements, the gain will be independent of frequency. However, if the effects of these reactive components are significant within the audio band, this linearity of frequency response will be modified, and this gives the designer considerable scope.

In order to do this kind of design work properly, it is very desirable to be able to work with complex numbers (ie, those containing the so-called imaginary value i or j , which is the square-root of -1). Doing the necessary

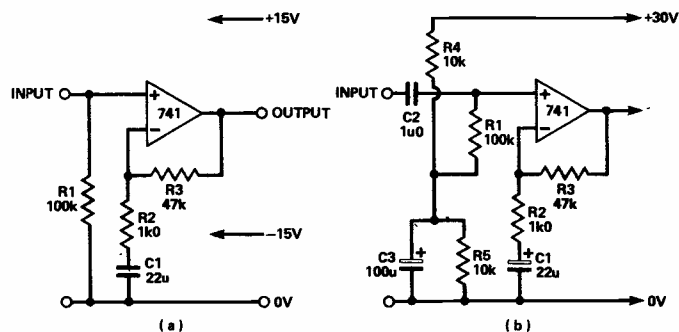


Fig. 3 Simple op-amp circuit. Note that in theory C1 should be unpolarsed, in practice an electrolytic is used. (b) is the circuit arrangement for use with a single-rail power supply, in all other respects it is equivalent to (a).

calculations with this type of equation is not really at all difficult, once some simple rules have been memorised, and this allows one to work out quite precisely how a circuit containing capacitors and inductors will operate, and gives both the actual gain and the phase shift. However, it is very hard to find text books which give a simple explanation and I propose, later on in this series, to give a non-mathematician's brief guide.

However, for the moment it is sufficient to remember that the -3 dB point (the frequency at which the gain is reduced to 71% of its flat response value) occurs in a stage such as that shown in 3a when the impedance of the capacitor C1 (given by $Z_c = 1/(2\pi fC)$) is equal in value to R2. Where the impedance of C1 is either very much less than R2 or very much greater than R2, the stage gain

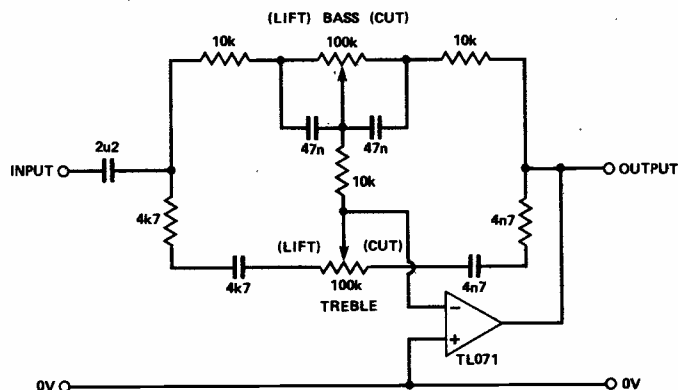


Fig. 4 'Baxandall' type tone-control circuit.

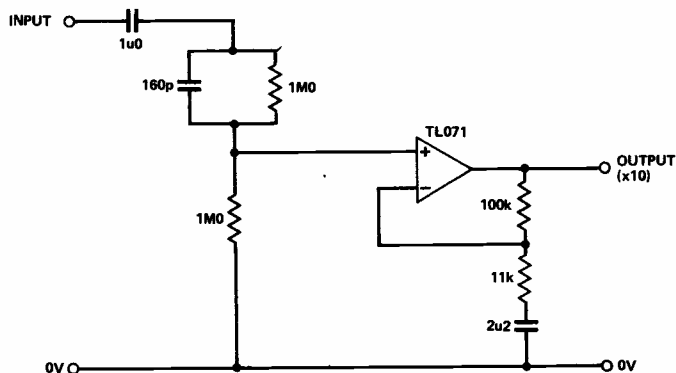


Fig. 5 Compensated ceramic pick-up input amplifier.

calculations can be simplified to $(R3 + R2)/R2$ or $(R3 + Z_c)/Z_c$. If one draws a graph and smoothly joins these three points, one will get a near-correct idea of the true way the circuit will behave.

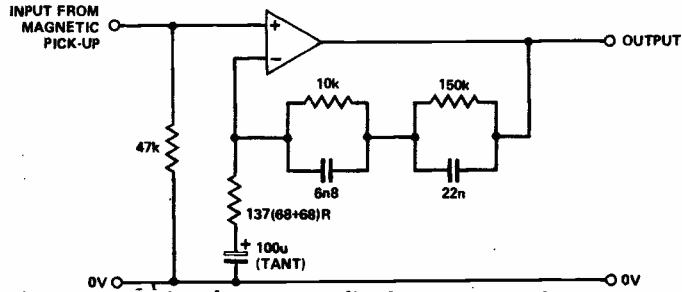


Fig. 6 Conventional RIAA-equalised magnetic pick-up amplifier stage (gain = 100 at 1 kHz).

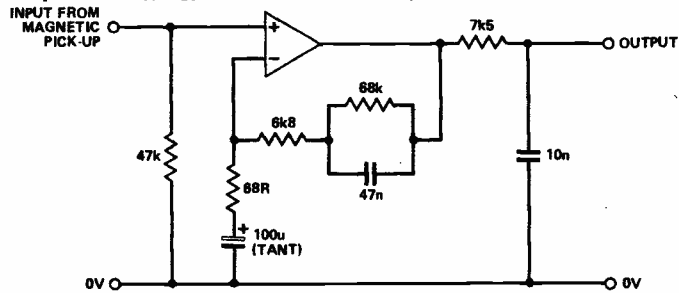


Fig. 7 Improved RIAA-equalised magnetic pick-up amplifier (stage gain = 100 at 1 kHz).

Thus the circuit varies its response according to frequency, and this is entirely due to the presence of the reactive components (more usually capacitors than inductors). Very many practical and useful designs are made possible: some examples are Fig. 4, which is a tone control circuit, and, as explained in the captions, Figs. 5, 6 and 7 which are pick-up amplifiers that accord with the RIAA specifications as shown in Fig. 9. The circuit arrangement shown in Fig. 7, in which the necessary double-step correction of the curve shown in Fig. 9, is done in two stages, is more accurate, particularly in respect of the sonically important transient performance, than the simpler, more commonly used arrangement of Fig. 6. The ceramic cartridge equalisation has a different requirement, since this is an amplitude — rather than velocity — sensitive device, and requires the type of replay curve shown in Fig. 10. Otherwise, when it is used with an adequately high input impedance to give a flat LF response, the reproduced sound is rather lacking in treble.

Combining bass cut and treble cut stages allows one to make local lift and local cut response circuits, such as those shown in Figs. 11a and 11b and whose performance is illustrated in Figs. 12a and 12b. Figs. 11a and b are really both the same circuit, but with the 'shelf' frequencies moved sideways.

I have illustrated all these later circuits using op-amps

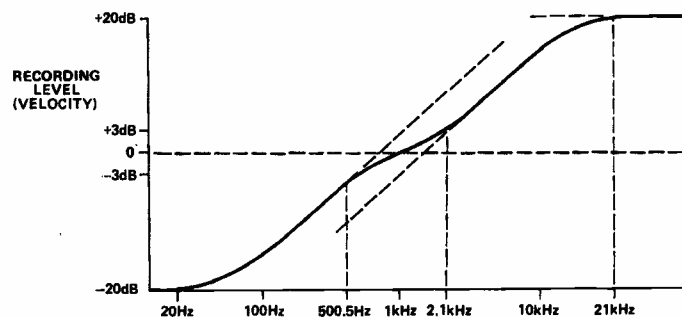


Fig. 8 Recording velocity characteristics employed in RIAA pre-emphasis convention for 33 and 45 RPM discs.

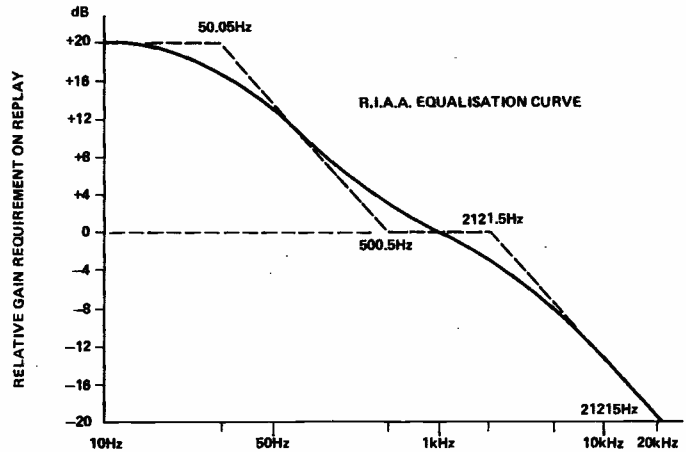


Fig. 9 Required replay curve for magnetic (velocity sensitive) pick-up cartridges.

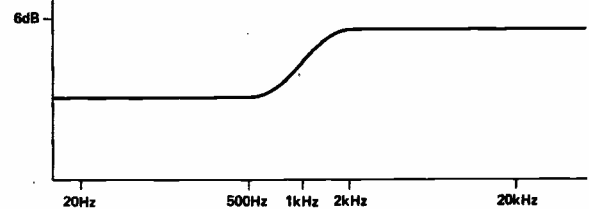


Fig. 10 Required replay curve for flat frequency-response output from ceramic (displacement sensitive) pick-up cartridges.

(a TL071 will work satisfactorily in all these). Let me confess that this was at least partly for convenience — it's much simpler (and easier to follow) if one can just show an amplifier as a triangle with three leads going to it. However, this also makes the point that circuit design (and circuit lay-out in the PCB) with op-amps is very much easier than with the equivalent discrete components (this is provided, of course, that the devices you are using have adequate performance in terms of output capability, distortion, noise, etc). But I still have the feeling that, for 'ultimate-fi', circuits using discrete components alone can be superior, if only because one can get the equivalent results with far fewer components (remember that a typical op-amp can contain the equivalent of 39 separate resistors, transistors, diodes, and capacitors, each with its own imperfections, all of them contributing to an accumulated total imperfection). Instinctively, I feel that the less one handles a hi-fi signal the better the end result.

Steeper cut circuits using two or more RC elements

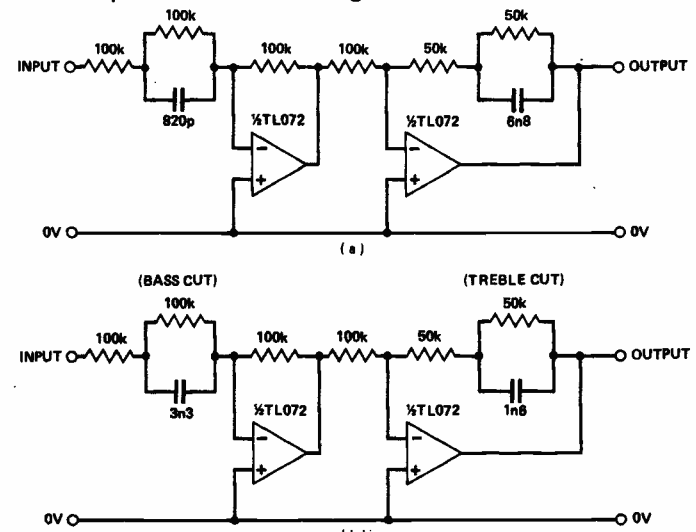


Fig. 11 Combined bass and treble cut circuits.

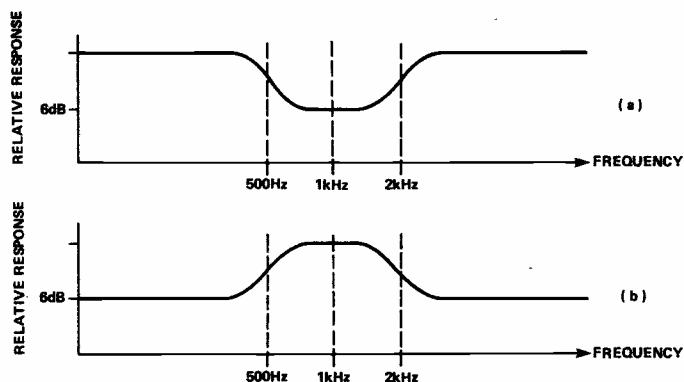


Fig. 12 Response curves for the two circuits in Fig. 11.

can be built, and if these RC groups are included in the feedback network we are now building active filters. These are great fun, but working out just what is going to happen requires rather more tricky maths, and a few useful dodges. Again, this is a topic I will defer until later. So, while I certainly use ICs in my own designs, and I accept that I will do so even more in the future as they get better and more versatile, nevertheless I do not see them displacing the circuits built up with separate transistors and resistors yet awhile. On the other hand, if one is making something like a car radio to which hi-fi standards are not really appropriate, or an FM tuner where the discrete component circuitry would be very cumbersome, *not* to use ICs would be truly ostrich-like behaviour.

Noise And Distortion — And Other Problems!

So far, I have largely pretended that we are living in an ideal world, where everything is as good as we would wish. Unfortunately, this isn't true, and the extent to which we will be successful in the field of audio design — as in any other — will depend on our ability to recognise the possible existence of defects, and to shape our designs, both on paper and as hardware, to avoid them. The problem, of course, is that it just isn't possible to optimise everything simultaneously, so what one ends up with must be a working compromise in which one has tried to assess what are the most important problems likely to affect one's listening pleasure, and to make sure that these are adequately dealt with.

This is, incidentally, one of the areas in which the DIY designer has a great advantage over the person who simply goes along to his hi-fi shop and hands over a wodge of pound notes for the latest black and chrome creation. This is because the commercial hi-fi is built to provide a good specification/price ratio, which will get a good review in the buyers' guides and ensure healthy sales. Unfortunately, no one really knows what makes hi-fi equipment, such as amplifiers, tuners and recorders, sound well; so, in the absence of any firm knowledge, a series of specifications relating to bandwidth, signal-to-noise ratio, power output, distortion and channel separation, have arisen — and these are the specs for which the commercial manufacturers seek to get good values. Whether the final thing sounds well cannot be so easily specified, but there are some areas, and I will look at these later, where something which measures less well does indeed sound better. So, if one is doing one's own thing one can design for sound rather than specifications.

Noise

If, in this term one includes all unwanted intrusions into the signal output, this consists of five main categories. These are: thermal noise, defective component noise,

radio breakthrough, impulse noise, and hum (from the AC mains power supply). There are also some other kinds of device noise with transistors (and ICs) which relate to the device operating conditions, and I have lumped these with thermal noise. I will leave this until last.

Hum: in any normal domestic environment, there is a possibility of the local (in the UK, 50 Hz) AC mains field intruding into the circuitry, even when this is battery operated. The only answer, in this case, is full screening of the lot, and care in the layout of connecting wires. Diecast metal boxes, such as those made by Eddystone and ITT, provide ideal housings for low-level and high-gain circuitry. Where one is powering equipment from the mains anyway, these problems multiply. Here one must make sure that one does not earth the equipment separately in more than one place (the all too familiar *earth loop* problem), one must make sure that the mains transformer has an adequately low external AC magnetic field, and that it is sited as faraway as possible from low signal level areas. Also, one needs to remember that the currents flowing in the transformer secondary, rectifier, and reservoir capacitor loop have very high peak values and will produce quite significant voltages across even small wiring resistances. Take care, therefore, to take off the DC supply from across the reservoir capacitor, including your 0V return! A further important point to watch is that there is no incipient instability in the circuitry of any amplifier, in that this will make it very prone to a hard rasping hum sound — similar to that given by two different earthing points on the mains DC supply.

Impulse noise: this has a lot in common with radio breakthrough, and is that annoying problem of clicks and bangs when other mains operated equipment in the same locality, lamps, fridges, central heating equipment, and so on, is switched on and off. This can be caused, partly, by the same things as radio breakthrough (see below), but is particularly exaggerated by incipient instability and unnecessarily wide gain bandwidth. If one had a moving coil head amp feeding the PU input, and if the head amp had a 10 MHz bandwidth, one would expect impulse noise problems. Thorough screening will also help here.

Radio breakthrough: for those who suffer from it, this is one of the most infuriating of problems — worse even than noisy neighbours. There is, unfortunately, no universal answer. In general, the problem arises because the various connecting leads act as aerials feeding signals into the amplifier, which are then rectified by slightly less than perfect plug-socket connections, transistors or IC semiconductors, or even by poor soldered joints. Decoupling those earth returns which are not directly connected to the chassis, via a non-inductive (eg, ceramic disc) capacitor will help, as will the connection of judicious 100 to 1000 pF ceramic capacitors between input transistor emitters and bases. Also, putting a few ferrite beads on likely live-side input leads can help. Alas, the trick which works in one case may be useless in another, so this requires a lot of experimentation, and not a little luck.

Defective component noise: in my early days as an electronics enthusiast, resistors which crackled, capacitors which spluttered, sizzled, and hissed, and valves which rang like a bell when one tapped them, were just part of life — and one tried, by replacement, to end up eventually with a good set, until yet another component 'went noisy'. Happily, things have changed for the better, and nowadays defective components are relatively rare, at least at normal signal levels, where an electrolytic capacitor installed with reversed polarity is likely to be the major noise culprit.

However, at low signal levels, such as at the phono input to an audio amp, this isn't so reliably the case, and

while proper selection of components for the application will help, for very low level use there is no real substitute for individual testing to guarantee good performance. Remember also that most electronic components don't like heat, so if one has had to spend a lot of time with the soldering iron bit close to the component in question — perhaps in taking it out of a PCB and in reinstalling it — it may not be as good afterwards as it was to begin with.

Thermal noise: this is the result of the random motions of electrons — the basic carrying elements — as a consequence of their being excited by heat. This noise component increases with absolute temperature ($^{\circ}\text{K} = 273 + ^{\circ}\text{C}$), and with the amount of resistance in the circuit. The formula for calculating this is

$$V = \sqrt{4k.T.\delta F.R}$$

where V is the mean AC output voltage, k is Boltzmann's constant (1.38×10^{-23}), T is the absolute temperature ($^{\circ}$ Kelvin), δF is the measurement bandwidth and R is the circuit resistance. It follows from this, immediately, that the larger the amount of resistance in the circuit (other things being equal) the worse the noise. A graph showing this relationship is given in Fig. 13.

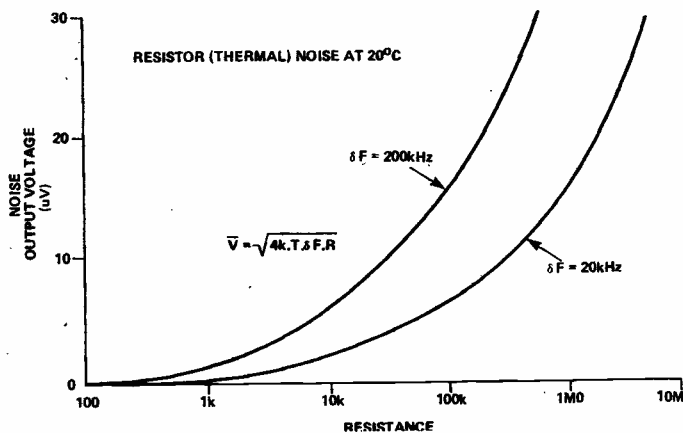


Fig. 13 Resistor thermal noise at 20°C.

Other sources of noise, related to circuit resistance, occur in junction and other transistors. These are those due to what is known as *base spreading resistance* (which decreases as the chip size and dopant concentration increase), and shot noise, which is a statistical problem and decreases, within limits, as the ratio of the total emitter current to the fundamental electronic current lumps, the electrons, increases. There is also *modulation* or *1/f* noise, which is proportional to the current and inversely proportional to frequency: this is very much dependent on the device used. Finally, there is reverse leakage noise, which worsens as the collector (or drain in the case of an FET) voltage increases.

To summarise this, the lowest noise in a transistor stage will be given by an optimally chosen device (in respect to type and performance), operated at the collector current which is best suited to its base and emitter circuit resistance and its chip size, at the lowest sensible temperature and operating voltage, and the lowest input impedance. Since all of the noise sources are bandwidth dependent, those which arise before a bandwidth limiting filter, such as an RIAA equalising stage, are less obtrusive than those which occur after this — other things being equal. Also, those noise sources which are inductive in character, such as a magnetic pick-up on the input to a preamplifier, have an impedance which increases with frequency, and will therefore give a worse noise level — since the ear sensitivity increases with frequency up to a few kHz — than a simple calculation based on its DC resistance would suggest.

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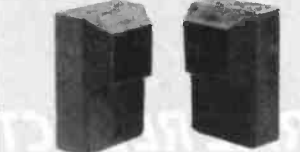
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MACHINE CODE PROGRAMMING

Machine code programming has two uses: firstly, as a way of getting your (normally BASIC-loving) microcomputer to go faster; and secondly, it's the only way to get a 'naked' microprocessor to do what you want. This latter use is what will primarily interest ETI readers. However, it's pretty difficult to learn on a bare micro, so in this short series, Bob Bennett will be showing us how it's done on a home computer, with some comments on using a microprocessor in the raw.

The best way to learn to programme using machine code is to have a go, after all, that was how you learned to programme in BASIC. But then, BASIC does bear some resemblance to everyday English, and machine code looks like . . . well . . . code, so how is it done? To answer that you need to have an insight into what is happening inside the computer — not a lot, just enough to make machine code programming clearer. In this magazine, starting in August of 1982, Owen Bishop wrote an excellent series on designing micro systems, and explained how all the pieces are put together to make a computer. I'll start with a short recap of some of the points relevant to machine code programming.

Deep In The Heart (of Texas?)

At the heart of any computer is a processor, and in most home computers it is a single chip. The Jupiter, Lynx, Spectrum and many others use a Z80 type, whilst the processor in the BBC, Vic 20, Apple, etc, is a type 6502. Each processor has its own instruction set, which is a repertoire of instructions the processor will obey, and each processor has a

register set, most of which can be used directly by the programmer. It is by the judicious use of the instruction set that the programmer manipulates the data in the registers to execute, in a controlled sequence, the various effects which constitute the desired aim of the overall programme.

CPUs differ quite a lot in both the sizes of their instruction repertoires and in the number of registers that they contain. We'll be looking at registers in a moment.

The two more common types of memory used in home computers are random access memory (RAM), and read only memory (ROM). Fundamentally they appear the same in general makeup, inasmuch as they both have a number of locations (called addresses) where data can be placed, but in ROM that data is sealed in and cannot be altered, hence read only. It is in the ROM where the designer has put the routines to control all the effects I mentioned earlier, such goodies as PRINT, PLOT, SCROLL, etc, in fact everything your computer can do. RAM is where the machine code programmer (that's you!) places the instructions (program) which the processor hopefully will obey. The designation random is a bit of a misnomer: there is nothing random in the way the memory is accessed, at least, not (we hope) in a computer!

Bits And Pieces

So what's the connection between RAM, ROM, registers and the processor? The answer is a bus. Not the number 8 to the office, but another name for a connecting wire, or, as is more usual in a computer, a group of wires (or tracks on a PCB). These wires carry information in the form of electrical signals, and it is the level of the voltages present on the bus which conveys the meaning of the signals. An acceptable high level can be taken to mean a 1, and an acceptable low level can signify a 0, which leads us to use binary notation on computing (convenient isn't it?).

If there are n wires making up a bus, then the total information on the bus can be represented as 2^n . Most home computers have eight-bit registers (where bit is a contraction of Binary digit), so the highest number this register can hold is 2^8-1 which is 255 if all the bits are 1s. These eight bits are known as a byte.

255 is not a very high number to play around with so it

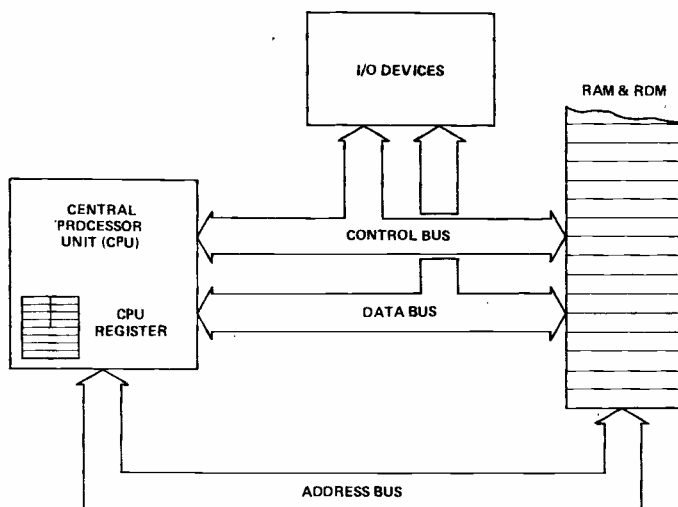


Fig. 1 Layout of a minimal computer.

is arranged that registers can be used in pairs, but only in certain combinations. This combination broadens our horizons somewhat because we can now use numbers up to 2^{16} which is equal to 65,536 decimal. The normal way to present data is one byte at a time, so our data bus usually has only eight wires. However, because we need a lot of memory, we use 16 wires on the address bus which allows up to 65,536 addresses, or locations to be used. This is known as 16K or 16 Kilobytes because it gets tedious writing

CONVERSIONS

CONVERSION OF HEXADECIMAL TO DECIMAL

A single hexadecimal register holds up to 256, and, as we do when counting in tens, we split this into a 16^1 figure and a 16^0 figure (as in tens and units). A register pair would hold figures for 16^3 , 16^2 , 16^1 and 16^0 .

hex	16^3	16^2	16^1	16^0
0	0	0	0	0
1	4096	256	16	1
2	8192	512	32	2
3	12288	768	48	3
4	16384	1024	64	4
5	20480	1280	80	5
6	24576	1536	96	6
7	28672	1792	112	7
8	32768	2048	128	8
9	36864	2304	144	9
A	40960	2560	160	10
B	45056	2716	176	11
C	49152	3072	192	12
D	53248	3328	208	13
E	57344	3584	224	14
F	61440	3840	240	15

Using the table: decimal 15 in a register pair = 000F whereas 240 decimal in a single register would = F0.

A0B0 hex = $40960 + 176 = 41136$ decimal

FEDC hex = $61440 + 3854 + 208 + 12 = 65514$

CONVERSION OF DECIMAL TO BINARY OR HEXADECIMAL

Conversion can be achieved in two ways, successive division or by spotting powers of two. Let's look at an example:

To convert 365 into binary by successive division goes as follows:

365 divided by 2 is 182 remainder 1
 182 divided by 2 is 91 remainder 0
 91 divided by 2 is 45 remainder 1
 45 divided by 2 is 22 remainder 1
 22 divided by 2 is 11 remainder 0
 11 divided by 2 is 5 remainder 1
 5 divided by 2 is 2 remainder 1
 2 divided by 2 is 1 remainder 0
 1 divided by 2 is 0 remainder 1

all successive divisions by 2 will yield the result 0 and the remainder 0.

The very first remainder we obtained the value of 2^0 , the next is 2^1 , the next is 2^2 , etc

So the binary for 365 is 0001 0110 1101 and the hex is 01 6D.

Spotting the powers of two would work as follows:

365 is over 256 (2^8) but under 512 (2^9) so the binary bit corresponding to 2^8 is 1

$365 - 256 = 109$

109 is less than 128, so the bit for 2^7 is 0

109 is greater than 64 so the bit for 2^6 is 1

$109 - 64 = 45$

45 is greater than 32 so the bit for 2^5 is 1

$45 - 32 = 13$

13 is less than 16, so the bit for 2^4 is 0

13 is greater than 8 so the bit for 2^3 is 1

$13 - 8 = 5$

5 is greater than 4 so the bit for 2^2 is 1

$5 - 4 = 1$

1 is less than 2 so the bit for 2^1 is 0

1 is equal to 1 so the bit for 2^0

We follow this through in the same way as before.

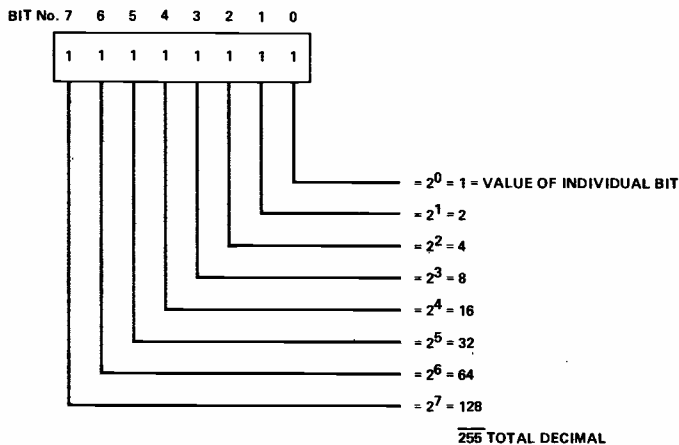


Fig. 2 The make-up of an eight-bit register.

out complicated binary numbers in decimal all the time. A K is 2^{10} , and this is equal to 1024 — it's the nearest convenient binary number to 1000, but note that a capital K is used to distinguish it from the decimal k (= 1000). We just have to hope that the context is such that any numbers are obviously not absolute temperatures (ie in degrees Kelvin or K)!

When you see advertisements extolling the virtues of home computers you will probably notice something along the lines of "16K ROM and 16K RAM". You will know that the ROM is for the routines that the designers have built into the machine. The start of the ROM area is usually (but not always!) address 0, so in the example given, it will extend up to address $16 \times 1024 - 1$, ie 16383 (the -1 is because we've started counting at 0 rather than 1 as is usual outside computers — think of a street with 16 houses, if the first is numbered 0, the last will be number 15).

Unfortunately, this doesn't leave the RAM entirely free for the user to place all his or her programs, data, etc, because the computer needs some space to use for its own internal housekeeping (it stores what are known as the systems variables). It is very important not to over-write or **corrupt** the areas that the computer needs for this purpose — doing so is a very effective way of bringing your micro to its knees (or whatever the micro equivalent of a knee is). However, even in the most modest of systems, there will be more than enough space left for a decent machine code program.

Do You Do Voodoo?

If you are a student of the occult you may have come across the word hex before (I believe it has something to do with casting a spell), but in computing circles it is a word that machine code buffs drop all over the place. Actually it is short for hexadecimal, where hexa is from the Greek pertaining to six, and decimal of course is all about tens, so putting them both together means we are counting using the base 16. Some people may believe that this is the Martian base for counting because they have sixteen fingers! Starting at zero (written as 0) we count up to 9 and then we go from A to F where A = 10 decimal and F = 15 decimal.

Note that we would write down 10 decimal as OA hex (or OAH), and 15 decimal is OF hex (or OFH): you must get used to the idea of writing hexadecimal numbers as two characters; for example, F on its own is meaningless whereas OF equals 15 decimal, and FO equals 240 decimal. FF hex equals 255 decimal which, if you remember, is the maximum that a register can hold, and also the number that eight bits would represent if they were all 1s which in turn represents one byte (see how it all fits in?), so two hex characters equal one byte. All this means that it is possible to write a machine code program in either binary, decimal or hexadecimal and still get the same result, but I think that you can discount using binary because it's far too cumber-

FEATURE : Machine Code Programming

some (although a knowledge of the binary system is essential for some applications as you will see).

To sum up so far: a machine code program is written to (or placed in) addresses in RAM one byte at a time, some bytes representing instructions, and some representing data. Registers, either singly or in pairs, are used to manipulate the instructions and data, and the processor sorts out all that little lot. According to the information in the programme, different routines in ROM are called into use to give different effects. This is a very simplified explanation, but essentially correct, and although I have only been talking about typical home computers, very much the same sort of process happens in larger computers, only on a much grander scale.

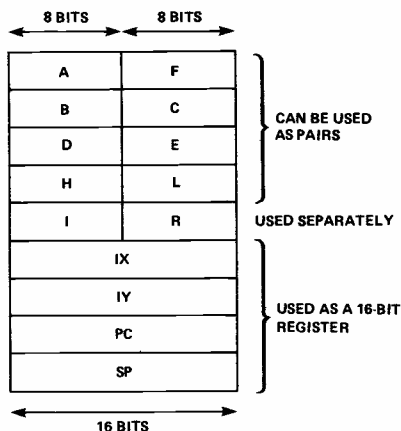


Fig. 3 The Z80 register set. Note that there is also an alternate set of registers A,B,C,D,E,F,H, L, usually referred to as A',B', etc.

I mentioned earlier that I would discuss registers in greater detail, so here we go. Using the Z80 set as a model (Fig. 3), the A register is historically called the accumulator because it was used to accumulate the results of computations. It is still a hard worked register, and there are certain operations that can only be carried out using the A register, but more of that later. The F register is the flags register alias the status register. This is so important to machine code programming that it warrants a section to itself. The B,C,D,E,H, and L registers are general purpose registers which are not found in a lot of CPUs.

When an input device requires the attention of the CPU it sends out a signal called an interrupt. What happens then depends on the CPU type, but usually an indicator signals the fact that an interrupt has occurred, and then the interrupt routine is entered. The Z80 has a rather unique way of dealing with an interrupt, however. Once an interrupt has been acknowledged, the device puts the low byte of an address onto the data bus. The high part of the address is in the I register, the two parts forming the address of a routine to handle the interrupt.

The R register is a simple counter (0 to 255) which is used to periodically refresh memory cells in RAM in order not to lose the contents. When a GOSUB is used in BASIC the computer uses a portion of RAM as a stack to store the address of the next instruction to be executed after meeting a RETURN. The stack is also used when pushing and popping (more later) to keep tabs on the addresses. It seems quite logical therefore to have a stack pointer to hold the address of the last item to be put onto the stack, this is the SP pair. The last registers in this set are the two used as a program counter (PC), the PC holds the address of the current instruction.

I have saved the two sets of register pairs IX and IY until now because not many CPUs have these sets. They are used

for indexed addressing which, very simply, is this, using IX as an example. The IX pair are made to hold the address of a table, where information relating to your program has been stored, this is known as the base address. When required the IX pair will meet instructions pertaining to their role in the program. These instructions are in two parts, the first part is a number, which is added to, or subtracted from the base address. This will point to an address in the table. The second part is an instruction relating to what will happen at that address, and this may, or may not influence what happens next in your program.

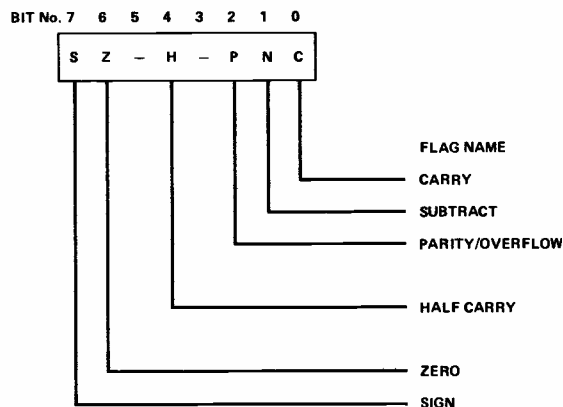


Fig. 4 The flags available in register F.

A Bit Of Flag Waving

As well as the general purpose registers, each processor will have a **flag**, or **status register**. These are constructed in exactly the same way as any other register, but the bits are used as indicators, or **flags**, to signal whether or not certain conditions have been met. The convention is that when a bit is set it is 1, and when reset, it is 0; when the condition has been met the flag is set, and reset otherwise.

Every micro I know of has a zero flag of some sort — one that is set when the contents of a particular register are zero. As an example, let's look at what is involved in the execution of a FOR-NEXT loop; something like this will be taking place: load a register with n (the loop count); do the task contained in the loop; decrement the count ($n = n - 1$); test the flag to see if the register is zero. If it is not, then go back and do the task again; if it is, go on to the next task. Note that both conditions of the flag can apply, and we program the computer to do one thing if the flag is set, another if it isn't.

The more usual flags are **zero**, **parity/overflow**, **sign**, **carry**, **half-carry**, **subtract**, and others may be **interrupt**, **decimal** and **break**. Whatever flags your processor uses, get to know them along with the instruction set. Any good computer handbook should give the instruction set, and any good library will have a computer section with a good selection of books on micros.

Other registers will include the **stack pointer** (SP), which may be a pair or a single register, which is used as a pointer to the stack area of memory. **Index registers** may come singly or in pairs, and are usually designated X and Y singly, and prefixed with I in pairs. As their name implies, these are used for indexing along tables of data. If you remember, a program is stored in a number of addresses, so a **program counter** (PC) is used as a pointer to these addresses. One last register: dynamic RAM will need refreshing (electrically) every now and again so that information isn't lost, so there is a **refresh register** (they think of everything). This list isn't exhaustive and don't worry if it isn't all completely clear what's going on. However, I hope that your appetite is whetted enough to probe further into your computer.

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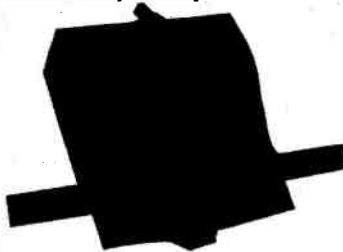
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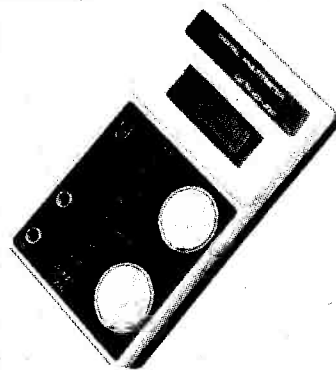


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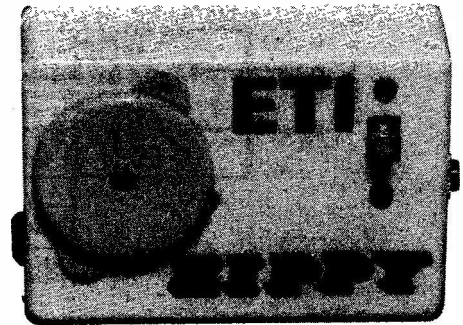
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SUPPLY PROTECTOR



Does that flickering of the house lights all too often indicate that your ZX has just had its memory corrupted? Here's a very simple remedy, designed by Phil Walker.

Designed primarily with the ZX81 home computer in mind but applicable to many others, this project aims to protect the program that you've just spent three hours correcting from short term mains failure or accidental supply disconnection. The sort of thing we mean is the temporary (or worse) dimming of the lights caused by lightning strikes on the grid lines or load switching at a sub-station. These effects usually only last a few tenths of a second but can cause your computer to forget itself and delete your program — resulting in instant frustration!

The solution is embodied in this project. What is needed is that the computer should be rapidly switched over to a standby battery. This need only be able to supply the current drawn for a few minutes until the normal supply is restored. The ETI Zippy does this and also sounds an alarm to tell you that something is wrong

The Circuit

The main part of the circuit consists of B1, D1 and IC1. B1 is a Nickel-Cadmium rechargeable battery with a capacity of 110 mAh at a voltage of around 8.4 volts. This means that when fully charged it should be able to supply a ZX81 for at least 6 minutes — longer if you do not have many extras plugged in. This will even give you time to save your program on tape (provided you have a battery powered tape recorder). D1 effects the switch-over from normal supply to Zippy's internal battery while IC1 recharges the battery while the mains is available. The rest of the circuit provides the audible warning signal from the piezo-electric sounder when the normal supply voltage drops too low.

It is probably a good idea to

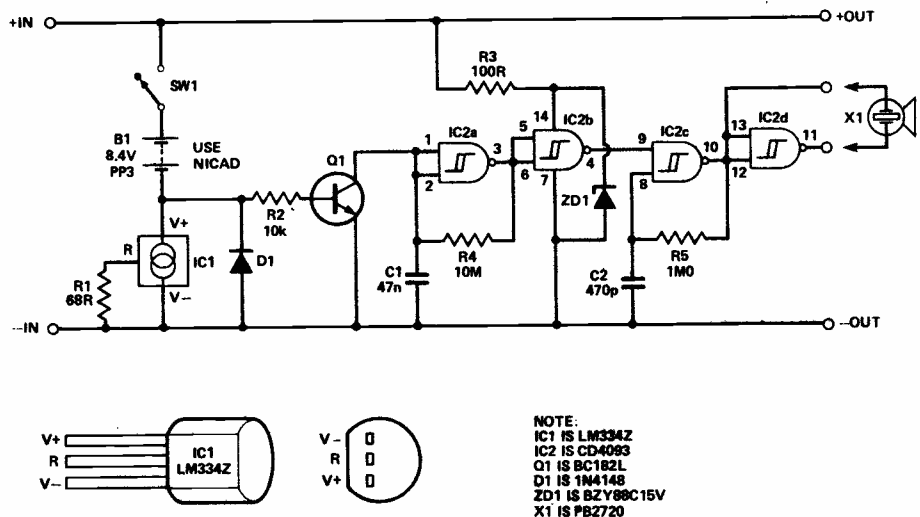


Fig. 1 Circuit diagram of Zippy

HOW IT WORKS

B1 is the main energy store with a capacity of 110 mAh and a terminal voltage of 8.4 volts. IC1 is a constant current device whose operating current is set at about 0.6 mA by R1. This level of current can be sent through the battery constantly with little degradation of performance and will keep it ready for use.

D1 blocks current flow from the power pack to the battery but will allow current to flow from the battery on to the supply lines if the power pack voltage drops below about 7.7 volts. This ensures that the supply lines never drop below this level. The internal regulator in a ZX81 needs about 6.5 to 7 volts minimum at its input pin to keep it working correctly.

While the input voltage from the power pack is more than a volt or so greater than the battery voltage, Q1 will be turned on by current flowing through R3. This will keep C1 discharged. This

will cause the outputs of IC2a and IC2c to stay high and IC2b and IC2d to stay low.

If the input voltage falls below this level, Q1 will turn off and allow C1 to be charged via R4 until it reaches the switching threshold of IC2a. The output of IC2a will now go low and C1 will be discharged via R4 until it reaches the lower switching threshold of IC2a whereupon IC2a output will go high again to repeat the cycle. While IC2a output is low the output of IC2b will be high. This enables a similar oscillator configured around IC2c. The frequency of IC2a oscillator is of the order of 2 Hz while that of IC2c is around 2 kHz; the resulting output from IC2c is bursts of 2 kHz which when applied to the piezo sounder make a slow bleep-bleep noise. IC2d is used to invert the output from IC2c and increase the signal voltage applied to the sounder.

charge the battery periodically so that you don't get caught out.

Construction

The project can be built into a small plastic box of the type made by Bicc-Vero (see Buylines). It is a tight fit in the one specified so some care must be taken when

siting the switch and input socket. The PCB is designed to fit along one side of the box with the battery along the other. Don't forget to cut the corners off the PCB where marked.

Assembly of the PCB is straightforward but care should be taken when fitting the diodes,

PROJECT : ZX PSU Protector

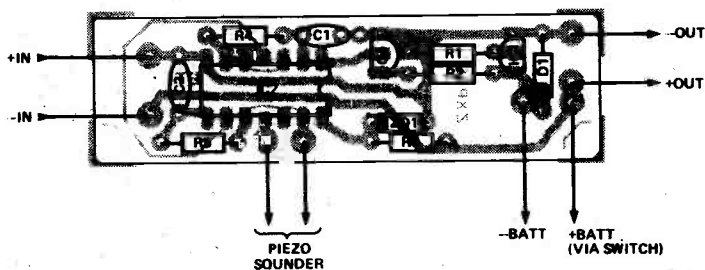


Fig. 2 Component overlay for the PCB

transistor and ICs that they are the right way round. Connect all the lead-out wires except those to the sounder before assembling the complete unit.

In our unit, the sounder was glued to the outside of the case and the wires taken inside through a small hole. Holes should also be cut for the switch, input socket and output wire. Make sure everything will fit before deciding where these holes will be.

For a ZX81, the input connector is a 3.5mm jack socket and the output wire is terminated in a matching plug (after assembly), but for your system these can be as required. Beware . . . not all power

supplies have the centre conductor positive, so check this before wiring up.

When everything is ready, put the PCB, switch and input socket in the case, thread the output lead out through the hole provided for it (you did cut one, didn't you?) fit a grommet if you want it to look nice, and wire up the sounder and other components as neatly as possible. Do not have the battery connected while you do this, as it has a very low impedance and can discharge with some violence. The PCB can be fastened in with a bit of sticky tape if you want but it cannot move about much in the limited space available.

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PARTS LIST

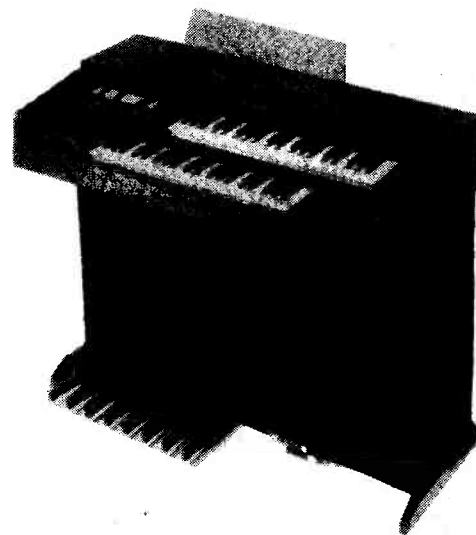
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R2	10k
R3	100R
R4	10M
R5	1M
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C1	47nF
C2	470pF
SEMICONDUCTORS	
IC1	LM334Z
IC2	CD4093
D1	1N4001 (Germanium or Schottky of 1A+ rating may improve performance)
ZD1	BZY88C15V (or similar 15V zener)
Q1	BC182L
MISCELLANEOUS	
X1	PB2720 piezo sounder
B1	8.4 V 110 mAh PP3 size Ni-Cd
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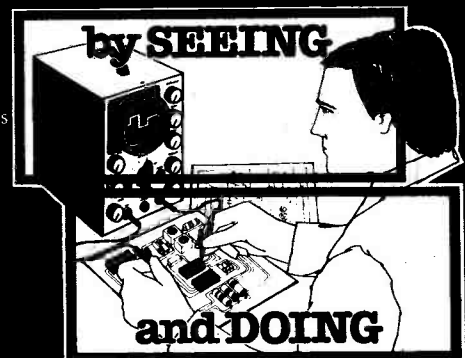
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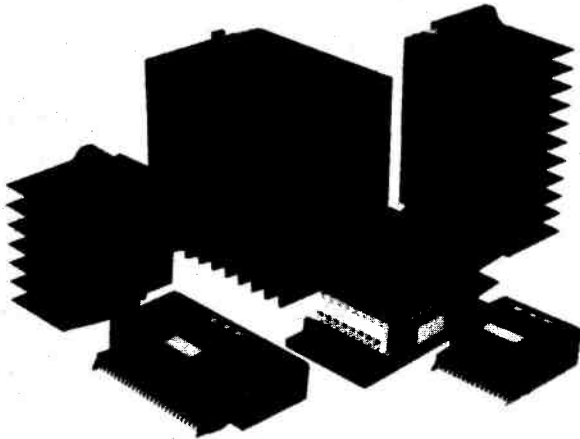
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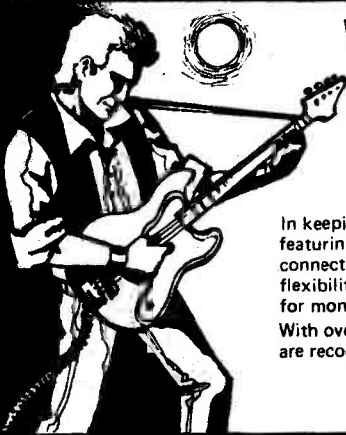
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HY6060	30 + 30	4-8	0.015%	<0.006%	± 25	120 x 78 x 40	420	£18.89
HY124	60	4	0.01%	<0.006%	± 26	120 x 78 x 40	410	£20.75
HY128	60	8	0.01%	<0.006%	± 35	120 x 78 x 40	410	£20.75
HY244	120	4	0.01%	<0.006%	± 35	120 x 78 x 50	520	£25.47
HY248	120	8	0.01%	<0.006%	± 50	120 x 78 x 50	520	£25.47
HY364	180	4	0.01%	<0.006%	± 45	120 x 78 x 100	1030	£38.41
HY368	180	8	0.01%	<0.006%	± 60	120 x 78 x 100	1030	£38.41

Protection: Full load line. Slew Rate: 15v/ μ s. Risetime: 5 μ s. S/N ratio: 100db. Frequency response (-3dB) 15Hz - 50KHz. Input sensitivity: 500mV rms. Input Impedance: 100K Ω . Damping factor: 100Hz >400.

PRE-AMP SYSTEMS

Module Number	Module	Functions	Current Required	Price inc. VAT
HY6	Mono pre amp	Mic/Mag. Cartridge/Tuner/Tape/Aux + Vol/Bass/Treble	10mA	£7.60
HY66	Stereo pre amp	Mic/Mag. Cartridge/Tuner/Tape/Aux + Vol/Bass/Treble/Balance	20mA	£14.32
HY73	Guitar pre amp	Two Guitar (Bass Lead) and Mic + separate Volume Bass Treble + Mix	20mA	£15.36
HY78	Stereo pre amp	As HY66 less tone controls	20mA	£14.20

Most pre-amp modules can be driven by the PSU driving the main power amp. A separate PSU 30 is available purely for pre amp modules if required for £5.47 (inc. VAT). Pre-amp and mixing modules in 18 different variations. Please send for details.

Mounting Boards

For ease of construction we recommend the B6 for modules HY6-HY13 £1.05 (inc. VAT) and the B66 for modules HY66-HY78 £1.29 (inc. VAT).

POWER SUPPLY UNITS (Incorporating our own toroidal transformers)

Model Number	For Use With	Price inc. VAT
PSU 21X	1 or 2 HY30	£11.93
PSU 41X	1 or 2 HY60, 1 x HY6060, 1 x HY124	£13.83
PSU 42X	1 x HY128	£15.90
PSU 43X	1 x MOS128	£16.70
PSU 51X	2 x HY128, 1 x HY244	£17.07

Model Number	For Use With	Price inc. VAT
PSU 52X	2 x HY124	£17.07
PSU 53X	2 x MOS128	£17.86
PSU 54X	1 x HY248	£17.86
PSU 55X	1 x MOS248	£19.52
PSU 71X	2 x HY244	£21.75

Model Number	For Use With	Price inc. VAT
PSU 72X	2 x HY248	£22.54
PSU 73X	1 x HY364	£22.54
PSU 74X	1 x HY368	£24.20
PSU 75X	2 x MOS248, 1 x MOS368	£24.20

Please note: X in part no. indicates primary voltage. Please insert "0" in place of X for 110V, "1" in place of X for 220V, and "2" in place of X for 240V.

MOSFET MODULES

Module Number	Output Power Watts rms	Load Impedance Ω	DISTORTION		Supply Voltage Typ	Size mm	WT gms	Price inc. VAT
			T.H.D. Typ at 1KHz	I.M.D. 60Hz/7KHz 4:1				
MOS 128	60	4-8	<0.005%	<0.006%	± 45	120 x 78 x 40	420	£30.41
MOS 248	120	4-8	<0.005%	<0.006%	± 55	120 x 78 x 80	850	£39.86
MOS 364	180	4	<0.005%	<0.006%	± 55	120 x 78 x 100	1025	£45.54

Protection: Able to cope with complex loads without the need for very special protection circuitry (fuses will suffice).

Slew rate: 20v/ μ s. Rise time: 3 μ s. S/N ratio: 100db

Frequency response (-3dB): 15Hz - 100KHz. Input sensitivity: 500mV rms. Input impedance: 100K Ω . Damping factor: 100Hz >400.

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C15

Mono Power Booster Amplifier to increase the output of your existing car radio or cassette player to a nominal 15 watts rms.

Very easy to use.

£9.14 (inc. VAT)

Robust construction.

Mounts anywhere in car.

Automatic switch on.

Output power maximum 22w peak into 4 Ω .

Frequency response (-3dB) 15Hz to 30KHz, T.H.D. 0.1% at 10w 1KHz

S/N ratio (DIN AUDIO) 80dB, Load Impedance 3 Ω .

Input Sensitivity and impedance (selectable) 700mV rms into 15K Ω 3V rms into 8 Ω .

Size 95 x 48 x 50mm, Weight 256 gms.

C155

Stereo version of C15.

£17.19 (inc. VAT)

Size 95 x 40 x 80. Weight 410 gms.

TOROIDALS

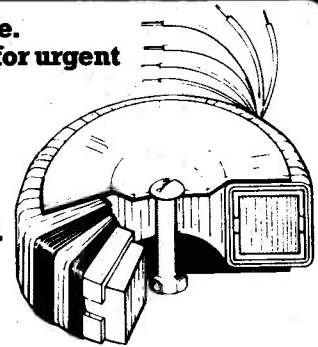
The toroidal transformer is now accepted as the standard in industry, overtaking the obsolete laminated type. Industry has been quick to recognise the advantages toroidals offer in size, weight, lower radiated field and, thanks to I.L.P., PRICE.

Our large standard range is complemented by our SPECIAL DESIGN section which can offer a prototype service within 7 DAYS together with a short lead time on quantity orders which can be programmed to your requirements with no price penalty.

***Gold service available.
21 days manufacture for urgent deliveries.**

***Orders despatched within 7 days of receipt for single or small quantity orders.**

***5 year no quibble guarantee.**



TYPE	SERIES No	SECONDARY Volts	RMS Current	PRICE	TYPE	SERIES No	SECONDARY Volts	RMS Current	PRICE	TYPE	SERIES No	SECONDARY Volts	RMS Current	PRICE
15 VA 62 x 34mm 0.35Kg Regulation 19%	0x010 0x011 0x012 0x013 0x014 0x015 0x016 0x017	6+6 9+9 12+12 15+15 18+18 22+22 25+25 30+30	1.25 0.83 0.63 0.50 0.42 0.34 0.30 0.25	£5.12 + p & p £0.78 + VAT £0.89 TOTAL £6.79	120 VA 90 x 40mm 1.2Kg Regulation 11%	4x010 4x011 4x012 4x013 4x014 4x015 4x016 4x017 4x018 4x028 4x029 4x030	6+6 9+9 12+12 15+15 18+18 22+22 25+25 30+30 35+35 110 220 240	10.00 6.66 5.00 4.00 3.33 2.72 2.40 2.00 1.71 1.09 0.54 0.50	£7.42 + p & p £1.72 + VAT £1.37 TOTAL £10.51	300 VA 110 x 50mm 2.6Kg Regulation 6%	7x013 7x014 7x015 7x016 7x017 7x018 7x026 7x025 7x033 7x028 7x029 7x030	15+15 18+18 22+22 25+25 30+30 35+35 40+40 45+45 50+50 110 220 240	10.00 8.33 6.82 6.00 5.00 4.28 3.75 3.33 3.00 2.72 1.36 1.25	£10.88 + p & p £2.05 + VAT £1.94 TOTAL £14.87
(encased in ABS plastic)														
30 VA 70 x 30mm 0.45Kg Regulation 18%	1x010 1x011 1x012 1x013 1x014 1x015 1x016 1x017	6+6 9+9 12+12 15+15 18+18 22+22 25+25 30+30	2.50 1.86 1.25 1.00 0.83 0.68 0.60 0.50	£5.49 + p & p £1.10 + VAT £0.99 TOTAL £7.58	160 VA 110 x 40mm 1.8Kg Regulation 8%	5x011 5x012 5x013 5x014 5x015 5x016 5x017 5x018 5x026 5x028 5x029 5x030	9+9 12+12 15+15 18+18 22+22 25+25 30+30 35+35 40+40 110 220 240	8.89 6.66 5.33 4.44 3.63 3.20 2.66 2.28 2.00 1.45 0.72 0.66	£8.43 + p & p £1.72 + VAT £1.52 TOTAL £11.67	500 VA 140 x 60mm 4Kg Regulation 4%	8x016 8x017 8x018 8x026 8x025 8x033 8x028 8x029 8x030	25+25 30+30 35+35 40+40 45+45 50+50 55+55 110 220 240	10.00 8.33 7.14 6.25 5.55 5.00 4.54 4.54 2.27 2.08	£14.38 + p & p £2.40 + VAT £2.52 TOTAL £19.30
50 VA 80 x 35mm 0.9Kg Regulation 13%	2x010 2x011 2x012 2x013 2x014 2x015 2x016 2x017 2x028 2x029 2x030	6+6 9+9 12+12 15+15 18+18 22+22 25+25 30+30 110 220 240	4.16 2.77 2.08 1.66 1.38 1.13 1.00 0.83 0.45 0.22 0.20	£6.13 + p & p £1.35 + VAT £1.12 TOTAL £8.60	225 VA 110 x 45mm 2.2Kg Regulation 7%	6x012 6x013 6x014 6x015 6x016 6x017 6x018 6x026 6x025 6x033 6x028 6x029 6x030	12+12 15+15 18+18 22+22 25+25 30+30 35+35 40+40 45+45 50+50 110 220 240	9.38 7.50 6.25 5.11 4.50 3.75 3.21 2.81 2.50 2.25 2.04 1.02 0.93	£9.81 + p & p £2.05 + VAT £1.78 TOTAL £13.64	625 VA 140 x 75mm 5Kg Regulation 4%	9x017 9x018 9x026 9x025 9x033 9x042 9x028 9x029 9x030	30+30 35+35 40+40 45+45 50+50 55+55 110 220 240	10.41 8.92 7.81 6.94 6.25 5.68 5.68 2.84 2.60	£17.12 + p & p £2.55 + VAT £2.95 TOTAL £22.62
80 VA 90 x 30mm 1Kg Regulation 12%	3x010 3x011 3x012 3x013 3x014 3x015 3x016 3x017 3x028 3x029 3x030	6+6 9+9 12+12 15+15 18+18 22+22 25+25 30+30 110 220 240	6.64 4.44 3.33 2.66 2.22 1.81 1.60 1.33 0.72 0.36 0.33	£6.66 + p & p £1.72 + VAT £1.26 TOTAL £9.64										

The benefits of ILP toroidal transformers

ILP toroidal transformers are only half the weight and height of their laminated equivalents, and are available with 110V, 220V or 240V primaries coded as follows:

IMPORTANT: Regulation — All voltages quoted are FULL LOAD. Please add regulation figure to secondary voltage to obtain off load voltage.

For 110V primary insert "0" in place of "X" in type number

For 220V primary (Europe) insert "1" in place of "X" in type number.

For 240V primary (UK) insert "2" in place of "X" in type number.

Also available at Electrovalue, Maplin

ALSO AVAILABLE
Sizes up to and including 5KVA are manufactured to order.

NEW PRODUCTS

HYBRID REGULATOR MODULES

The HR314 and HR614 regulated power supplies provide a constant 13.8 volt d.c. output at up to 3 Amp or 6 Amp respectively. The modules are encapsulated to an integral heatsink and are fully short circuit protected making them suitable for home or bench running of CB, car stereos or any 12 volt d.c. equipment required for many hobby or professional applications.

HR314 £10.23 inc. VAT

HR614 £18.51 inc. VAT

TECHNICAL SPECIFICATIONS

MODULE	HR314	HR614
Output Voltage	+13.8v ±5%	+13.8v ±5%
Output Current	Up to 3A	Up to 6A
Current limit (nominal)	3.5A approx	7A approx
Maximum Input Voltage	+30v	+30v
Minimum Input Voltage	+16v	+16v
Maximum Input Voltage for nominal output current	+20v	+20v
Maximum output current at 30v input	1.8A approx	3.5A approx
Output ripple (100Hz) - See Note 1	<10mV rms	<10mV rms
Size in mm.	76 x 68 x 40 high	120 x 78 x 40 high

POWER SUPPLY UNITS: comprising toroidal transformer plus 90 x 50 x 55 mm high printed circuit board containing smoothing and rectification

PSU31X Suitable for running one HR314 at full rated current. £13.17 inc. VAT
PSU56X Suitable for running one HR614 at full rated current. £19.13 inc. VAT

For 110v operation insert 0 in place of X — brown primary leads.
220v (Europe) operation insert 1 in place of X — pink primary leads.
240v (UK) operation insert 2 in place of X — orange primary leads.

For mail order please make your crossed cheques or postal orders payable to ILP Electronics Ltd. Barclaycard/Access welcome. Trade orders standard terms.



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CURVING ELECTRONS

Magnetic fields can be mind-bending when you try to understand what's going on. They can also bend the paths of poor little innocent electrons, too. John Dance shows us a practical import of this phenomenon.

Discovered as long ago as 28th October, 1879 by Edwin Hall of the John Hopkins University, Baltimore, the Hall effect found few applications until high quality semi-conductor materials became available since it is so small that it is difficult to detect in metals. Hall found that if a magnetic field is applied to a current carrying conductor at right angles to the direction of the current flow, a potential difference appears across the material in a direction which is at right angles to both the direction of the current flow and to that of the magnetic field.

In Fig. 1(a), the potential applied between the two electrodes causes an electric current to flow through the material. If this material is homogeneous and no magnetic field is present, the current flow through it is of uniform density. In the case of the P-type semiconductor material shown, current is effectively carried by the majority hole carriers which behave as positive charges, and these move in the same direction as the conventional current flow in the external wires.

If a magnetic field is now applied so that its direction is into the paper, Fleming's left-hand motor rule indicates that the moving holes will experience a force towards the left and will tend to curve in this direction, as shown in Fig. 1(b). As holes cannot flow out of the left-hand face of the block of P-type material, some positive charges will accumulate there. Similarly, negative charges will accumulate on the right-hand face of the block of material since no holes can flow into this face.

The electric field created by these charges tends to repel the holes from the positively charged left-hand face

towards the negatively charged right-hand face. The field increases until the positive charges are again moving uniformly across the block of semiconductor material as shown in Fig. 1(c). Any tendency to move to the left will increase the electric field, causing the charges to move directly across the block of material so that the balance is accurately restored. A pair of Hall electrodes placed in the position shown in Fig. 1(c) can be used to detect the Hall voltage produced in this way.

In the case of N-type semiconductor materials in which the majority carriers are electrons, the flow is in the opposite direction to that of the conventional current in the external wires. The left-hand rule again shows that the charge carrier movement is towards the left, but in this case the negative charge carriers build up a negative charge on the left-hand side and a positive charge on the right-hand side. Thus we can use the Hall effect to distinguish between N and P-type materials by detecting the polarity of the Hall effect voltage produced.

In most metals one obtains a Hall effect voltage with the same polarity as in an N-type semiconductor material, since conduction is by means of electrons. However, the Hall voltage is much smaller than in semiconductor materials and a few metals, such as zinc, produce a Hall voltage of the opposite polarity; in such metals the interaction of the moving electrons with fixed positive ions results in the current being effectively carried by holes. Intrinsic (pure) semiconductor materials show a small Hall effect; although the numbers of electrons and holes per unit volume are approximately equal, the electrons are more mobile, and the overall behaviour is normally like that of an N-type material.

The Hall effect in semiconductor materials produces a much larger Hall voltage than in metals because the number of charge carriers per unit volume is far smaller. The Hall voltage, V_H , is given by the equation:

$$V_H = \frac{BI}{Net}$$

where B is the magnetic flux density
 I is the current flowing through the specimen
 N is the number of charge carriers per unit volume
 e is the charge of an electron (1.6×10^{-19} coulombs)
 t is the thickness of the specimen.
 If one considers a piece of copper of thickness 1 mm carry-

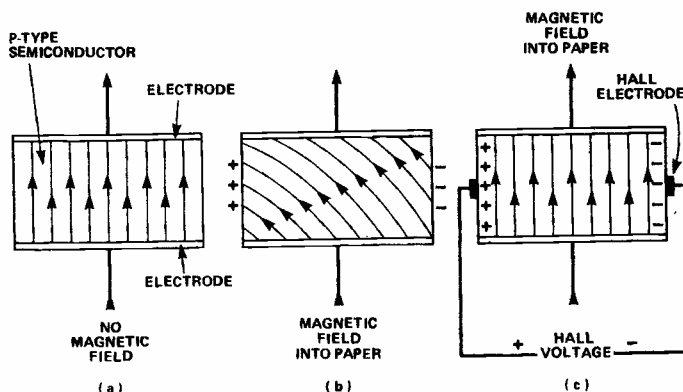


Fig. 1 (a), (b) and (c) The Hall effect illustrated by hole flow.

ing a current of 1A in a magnetic field of 1 Tesla (10,000 Gauss), V_H works out as a mere 62.5 nV, since N is about 10^{29} electrons per m^3 for copper. It is extremely difficult to measure 60 nV in such a circuit. In silicon, however, N may be 10,000 times smaller, so under the same conditions one obtains a V_H value of 625 μV which is a much more reasonable voltage for measurement. Hall first detected the effect using a thin gold foil.

The Hall effect has been widely used in materials science research where it enables information to be obtained about the charge carriers. When indium antimonide semiconductor Hall cells became available, they were used for the measurement of magnetic fields, but indium arsenide produces a Hall cell with about one-tenth of the temperature coefficient of indium antimonide, although its Hall output voltage is lower. Hall cells can be used as multiplying devices, for example in wattmeters, where a voltage is used to generate a proportional current through a Hall cell using a series resistor, while the load current passes through coils which generate the magnetic field in which the Hall cell is placed. Thus the output is proportional to the product of the voltage and current.

Other important applications of discrete Hall devices include their use in brushless DC motors in which the conventional brush and commutator system is replaced by Hall effect devices and suitable amplifiers. The use of such brushless motors avoids the inconvenience of brush replacement and improves reliability, but the elimination of sparking at the brushes is perhaps the most important advantage where low electrical noise is important (such as in high quality tape recording equipment).

Hitachi developed a Hall effect tape replay head in 1977 which is stated to have a high signal-to-noise ratio and excellent transient response (since it is non-inductive). A thin film of indium antimonide is employed, the output being determined only by the magnetic flux present so that a constant response is obtained at the lower frequencies — even down to zero frequency.

Monolithic Hall Devices

Silicon Hall effect devices have the great advantage that other circuitry can be integrated on the same silicon chip using normal IC production processes. Unfortunately the Hall voltage from silicon cells is about a hundred times smaller than that of Hall devices made from indium arsenide, but the temperature stability of silicon devices is far better and the small output levels can be amplified by on-chip components. Typical Hall voltage outputs from silicon cells are in the millivolt to tens of mV region, depending on the operating conditions.

Although most Hall effect devices are used in switching circuits, there are plenty of applications for linear

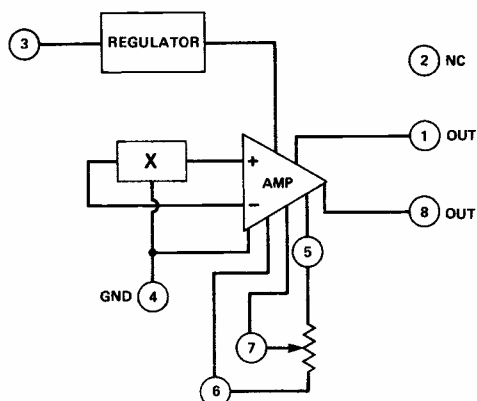


Fig. 2 Block diagram of a monolithic linear Hall effect device.

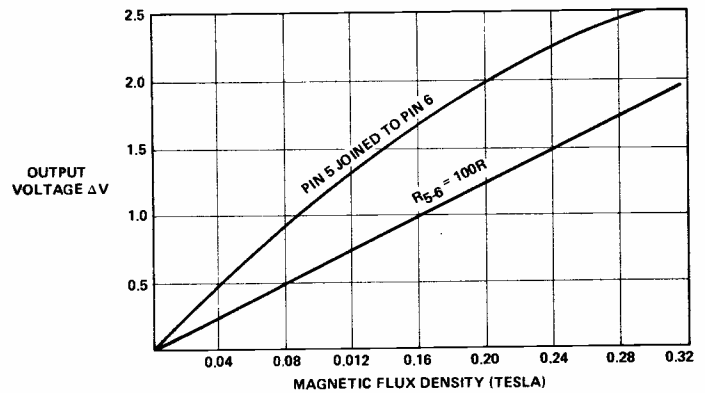


Fig. 3 Output voltage versus magnetic flux density for a UGN-3501M.

Hall devices. The basic internal circuit of the Sprague UGN-3501M linear device is shown in Fig. 2; it can be seen that the small output from the Hall cell itself is amplified by an op-amp. Offset output nulling facilities are included in this eight-pin DIL device, but not in the UGN-3501T which has only 3 connections. The UGN-3501T operates on from 8 to 12 V and the UGN-3501M from 8 to 16 V power supplies. The output voltage from a UGN-3501M device at various values of magnetic flux density with a 12 V supply, a 10 k load and two different values of resistor between pins 5 and 6 are shown in Fig. 3. The frequency response of these devices extends to about 25 kHz (-3dB). The sensitivity of the UGN-3501T is roughly twice that of the UGN-3501M.

Fig. 4(a) shows an application of the UGN-3501T as a ferrous metal detector and Fig. 4(b) is the circuit used. The pole of the magnet is fixed in contact with the Hall device and the output falls by 20 mV peak as the 25 mm steel ball rolls above the sensing device. This signal is amplified by a 741 device and drives the 2N5812 to conduction so that 0.5 A passes through the load. The low frequency response may be controlled by changing the value of the 22 μF coupling capacitor and high frequency attenuation may be introduced by using a small capacitor to shunt the feedback resistor of the 741.

By attaching the opposite pole of the magnet to the

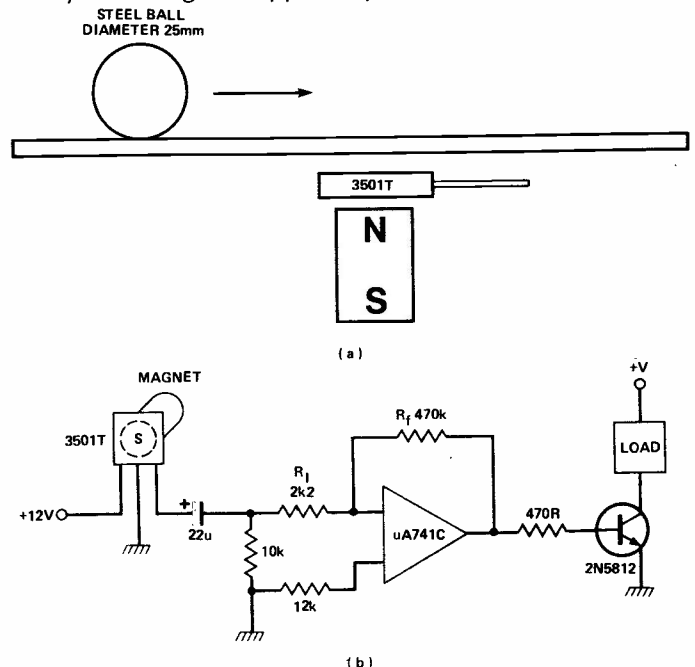


Fig. 4 (a) and (b) A ferrous metal detector using the UGN-3501T.

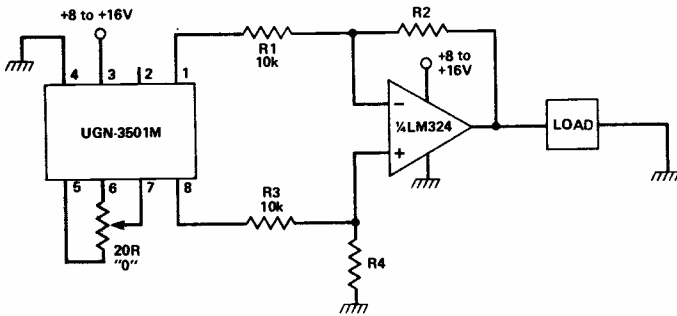


Fig. 5 A Hall effect switch using the UGN-3501M.

Hall device, it can be made to sense the *absence* of ferromagnetic material rather than its *presence*.

Fig. 5 shows the use of a LM324 operational amplifier to supply a voltage gain and to transform the differential output of a UGN-3501M into a single-ended output so that the circuit can drive a load which has one side earthed. The LM324 can be operated from a single power supply provided that the output does not swing below 0V. The connections shown are suitable for the detection of the field from a south pole, but if that from a north pole is to be detected, pins 1 and 8 should be reversed.

Another application for linear devices is in flux-meters, but calibration will be required. A typical UGN-3501M provides a differential output of about 1.4 mV in a 0.1 T field. The response is quite linear to 0.1 T, but the useful linear range can be extended to 0.3 T if a resistor of about 47 R is placed between pins 5 and 6 (see Fig. 2).

Linear devices can also be employed in current measurement applications. The device may be placed in the gap of a toroid and the current passed through a coil on the toroid. This may be used for overload detection in electric motors, current limiting, etc.

Siemens have recently introduced a KSY 10 linear Hall effect position sensor in which a gallium arsenide (GaAs) substrate is employed. This device is unique in that it is manufactured by an ion implantation planar technique which produces a doped layer only 0.3 μm in thickness; the use of this thin layer enables a sensitivity of 200 ± 30 V/AT to be obtained with a temperature coefficient of only about $\pm 5 \times 10^{-4}$ per degree K. For example, it will provide a Hall output of about 200 mV with a 5 mA control current in a field of flux density 0.2 T. The sensitivity can be selected in the range 30 to 300 V/AT by choosing the appropriate ion doping level during manufacture. The two Hall voltage output connections and the two control current connections are interchangeable, since the active sub-regions are symmetrical.

The output from the KSY 10 device is proportional to the effective magnetic field and to the control current passing through the device. The sensor is only 1 mm deep, so it can easily be positioned in the magnet yoke of current converters for current measurements. The active area itself is a mere 0.2 mm by 0.2 mm and lies 0.35 mm behind the front of its mini-plastic case. The device is very suitable for determining the position or speed of toothed gears or of rack and pinion mechanisms. The wide band gap of the gallium arsenide material used enables this device to be used at temperatures of up to 150°C, so applications in the engine compartment of motor vehicles are envisaged and it may also be used in brushless DC motors.

It is interesting to note that Yoshito Takehana's Group of the Electronic Devices Development Division of the Sony Corporation of Tokyo has developed a very sensitive silicon Hall effect sensor inside a special transistor. The output terminals of this magnetic sensor are in the reverse

biased depletion layer; a magnetic field perpendicular to the flow of the charge carriers between the base and collector terminals will produce an output of about 85 V/cm at a flux density of 0.1 T. If such a linear device is successfully developed to the production stage, a much wider field of application may be opened to Hall effect sensor devices at some future date.

Switching Devices

Switching or digital Hall effect monolithic devices are especially easy to use and are finding many applications in keyboards, in vehicle circuits, in toys and in any applications where movement must be converted into an electrical digital type of signal.

The basic circuit of a typical Hall effect switching device is shown in Fig. 6. An on-chip regulator is usually incorporated in the device, since this is necessary to produce a constantly repeatable performance, especially in automobile applications where the supply voltage can vary over a wide range. The regulator supplies a constant current through the Hall cell (shown by a X in Fig. 6) and the two connections which supply the Hall output voltage feed the inverting and non-inverting inputs of a comparator device which in turn drives a Schmitt trigger circuit and an output stage.

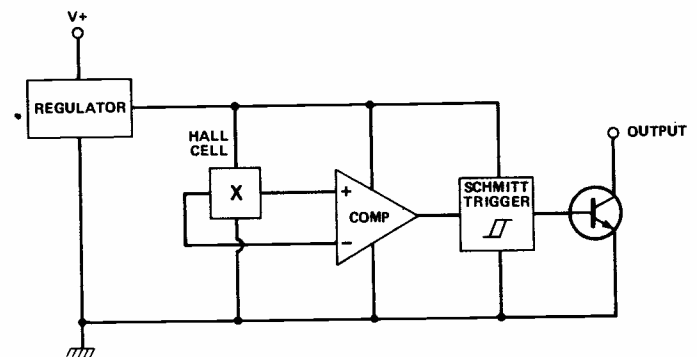
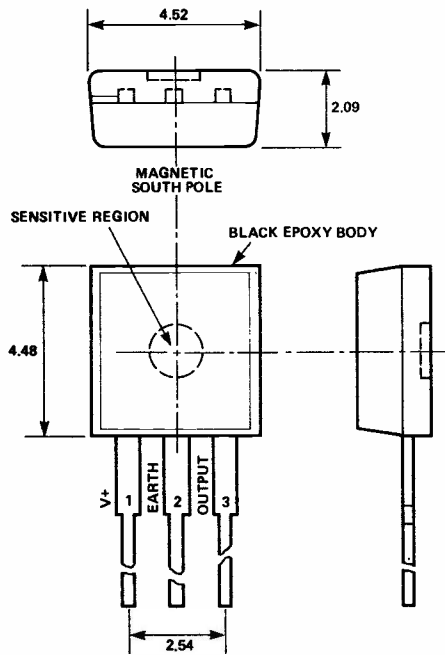


Fig. 6 Block diagram of a monolithic switching Hall effect device.

When the magnetic flux density in the Hall cell changes, the Hall voltage from this cell will change so that the comparator will switch the state of the Schmitt trigger circuit. A suitable amount of hysteresis is built into the circuit so that if a small increase in the magnetic flux density causes the output to switch into its other state, an appreciably larger decrease in the flux density will be required to cause the circuit to switch back to its former state. This prevents repeated rapid switching between the two states for very small changes in the flux density.

The Sprague UGN-3019T device (formerly coded ULN-3006T), is an economical product very suitable in most applications for the experimenter. This is encapsulated in the T-type package shown in Fig. 7, the Hall element itself being at the centre of one face of the device as indicated. As Hall effect devices are used in conjunction with a magnetic field, it is obviously important that the package used should allow the device to be easily orientated with respect to the field and to be easily mounted. The type of package shown is, in the opinion of the writer, usually more convenient for magnetic field sensing than the dual-in-line packages sometimes used for Hall Effect sensors. The UGN-3201M (formerly designated ULN-3006M) is very similar to the UGN-3019T, but is mounted in an 8-pin dual-in-line package.

The UGN-3019T may be used in the basic circuit of Fig. 8. In the absence of any magnetic field, the internal output transistor is cut off and passes only a very small col-



ALL DIMENSIONS IN MILLIMETRES

Fig. 7 The UGN-3019T — "T" type package.

lector current (typically, $1 \mu\text{A}$, maximum $20 \mu\text{A}$). The output voltage is therefore 'high' and has a value which is almost equal to the positive supply voltage; this supply voltage may have any value from +5 V to +16 V with an absolute maximum of +20 V (above which the device may be damaged).

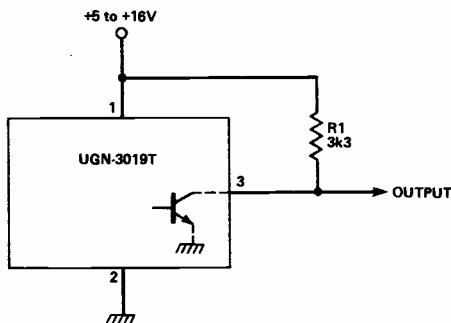


Fig. 8 A Hall effect switch using the UGN-3019T.

If a magnetic field of adequate flux density and of the correct polarity is now applied perpendicular to the face of the device, the internal Hall cell provides a voltage to the comparator of Fig. 6 which switches the Schmitt trigger so that the output transistor in the device conducts. The output falls to its low state with a typical value of +0.15 V and a maximum value of +0.4 V. The UGN-3019T can sink a current to its output of up to 15 mA, so the load resistor R1 of Fig. 8 must be chosen so that not more than 15 mA will flow into pin 3 with the particular value of positive supply voltage used.

The writer found that a UGN-3019T would switch to its low voltage output state when a small bar magnet was brought within about 3 mm of the centre of the body of the device. Owing to the built-in hysteresis in the internal circuit of the device, it did not revert back to the 'high' output state until the bar magnet was withdrawn to a distance of over 10 mm. The hysteresis characteristics of the UGN-3019T are shown in Fig. 9. A typical device switches to the 'low' output state at a field of 0.05 T and all devices are certain to switch at a field of 0.075 T at the centre of their face. A typical device reverts to the 'high'

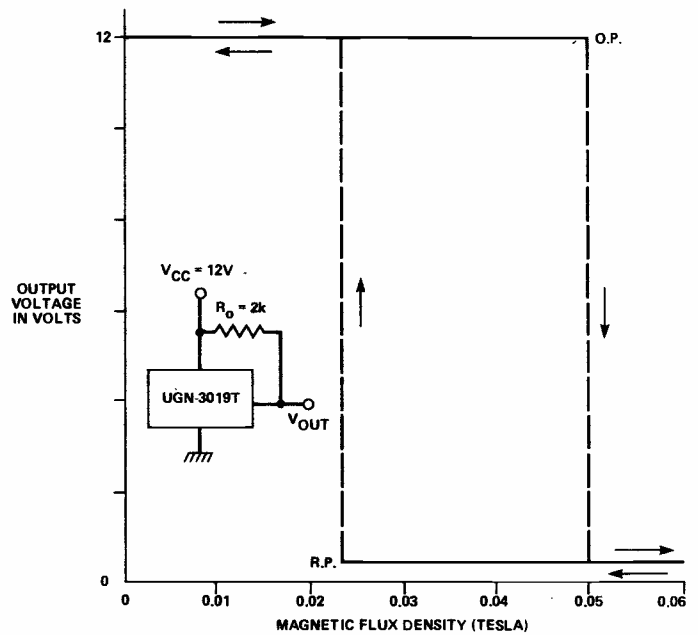


Fig. 9 Output voltage versus magnetic flux density for a UGN-3019T.

output at 0.0225 T and all devices at a value not less than 0.01 T. The device is unaffected by small stray magnetic fields from any transformers, relays, etc. which may be near to it.

UGN-3019T circuits are unaffected by the application of a field of the opposite polarity to that required to switch the output to the low voltage state. If the field is too weak to cause switching to the low output state, an improvement in the sensitivity can be obtained by placing a piece of iron or other ferromagnetic material on the far side of the device from the magnet as close to the device as possible. A greater increase in sensitivity can be obtained if the device is placed between two magnets with opposite poles on each side of the device. It is important that the magnet should be moved on a line directly towards the centre of the device, since a displacement of about 3 mm from the centre line can more than double the required flux density.

The UGN-3019T requires a supply current of about 7 mA (maximum 9 mA) with a 5 V supply and about 12 mA (maximum 16 mA) with a 12 V supply. A particular advantage of Hall effect switching devices over mechanical contact switching is their high speed of operation, the rise and fall times being measured in nanoseconds with operating speeds of up to about 100kHz. The output pulses are 'clean' without the 'bouncing' which is characteristic of mechanical contacts. Monolithic Hall effect devices are comparable in price to reed switches.

Using Hall Switches

An important use of Hall effect switching devices is for the detection and measurement of rotation. Many types of mechanical system arrangement are possible. In the 'slide by' mode one or more small magnets are mounted on a spinning disc and these magnets pass close to the face of a Hall effect IC. Each time a magnet passes the device, the circuit switches first to its low output voltage state and then back to the high output voltage state as the magnet moves away from the device.

An alternative system is the 'vane switch' technique, in which soft iron vanes attached to the rotating metal disc pass between the magnet and the Hall device. Each time a vane passes through this gap, the magnetic flux no longer

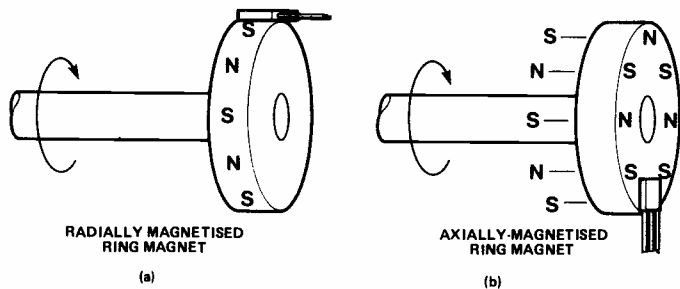


Fig. 10 (a) and (b) Ring magnet revolution indicators.

reaches the Hall device owing to the shielding effect of the iron in the vane, and the Hall circuit switches back to its high output voltage state.

Fig. 10(a) shows a system for detecting rotation which uses a radially-magnetised ring magnet. Suitable inexpensive ring magnets for use with either type of system are readily available. Up to eight pulses per revolution per 10 mm diameter of the magnet disc are possible, so 80 pulses per revolution can be obtained from a 100 mm diameter disc. These two arrangements have been designed for the UGN-3030T device; this is similar to the other devices discussed except that switching to the low voltage state occurs at a typical flux density of 0.016 T (maximum 0.025 T) and return to the high voltage state at 0.011 T (minimum -0.025 T). The power supply current required is only about half that needed for the UGN-3019T. It should be noted that to ensure switching of the UGN-3030T back to the high voltage state, a field of the opposite polarity is required of flux density -0.025 T; this is provided by the use of alternate polarity magnetic poles in the ring magnets of Figs. 10(a) and 10(b).

Rotational systems as described above have a very wide range of use in engines and machinery. One that many readers will have first-hand experience of is in car ignition circuits, where the contact breakers are replaced. This leads to the ignition timing being a once-and-for-all setting, as there is no wear, and this can only help improve fuel economy and lower exhaust pollution.

The same sort of sensor head can be used for measuring rotational speeds and counting the number of revolutions. In this case possibly the best course to take with the electronics is to have a pulse-generating circuit after the Hall effect device, so that the pulses can either be counted or fed to an analogue meter (to get a rate of revolution indication).

When fitted to a vehicle wheel such a system could have a further important use, namely in an anti-skid braking system. In this the electronics would detect when the wheel was not turning while the car was still moving. The system would then momentarily reduce the brake

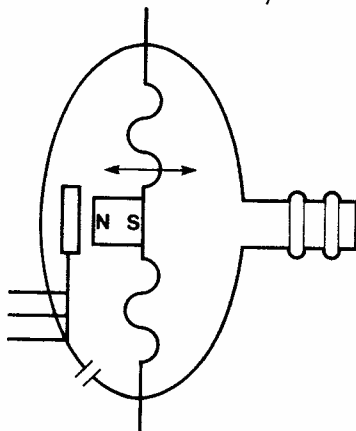


Fig. 11 A Hall effect pressure switch.

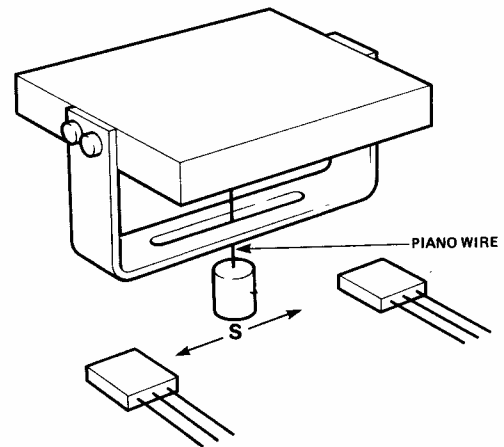


Fig. 12 An acceleration sensing system.

force to the wheel so that it would turn again and control would be restored, after which full braking force would be restored. This has the effect of pumping the brakes — but be warned, it's not easy to construct such a system, and we strongly recommend not trying!

Hall effect devices can be used to detect linear motion; Fig. 11 shows a simple pressure switch. Coupled with a push-button, this sort of arrangement is common in keyboard switches. A similar application is as an acceleration detector, and Fig. 13 shows how this can be done. In this, acceleration forwards or backwards causes the magnet to move nearer to one of the two Hall devices. Conversely, the tilt sensor in Fig. 13 works by detecting when the magnet moves away from directly above the Hall device.

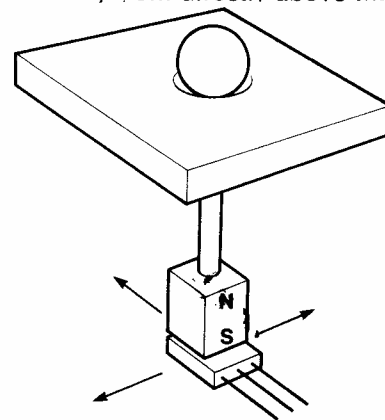


Fig. 13 A Hall effect tilt sensor.

As transducers go, Hall Effect devices can give a relatively large switching capability, being capable of sinking ample current to interface directly with TTL. Fig. 14 shows a suitable circuit for interfacing to CMOS devices.

Fig. 15 shows a handy buffer circuit that can be used to drive larger loads, such as a 12 V relay coil. In Fig. 14, when the magnetic field is strong enough, the output from the Hall device will be low and the transistor will be off. Hall devices can drive reed relays directly provided that they do not pass too much current, and provided that a transient suppressing diode is connected across the coil to prevent the back-EMF from destroying the Hall device (the diode cathode should go to the positive terminal of the relay coil).

If a Hall device such as the UGN-3030T is required to control a triac such as the RCA 40669 which can handle up to 8 A RMS, a transistor amplifier stage is required between the Hall device and the triac as shown in Fig. 16. When the Hall device conducts, a current of 9 mA flows from the base of the PNP 2N5811 transistor which in turn

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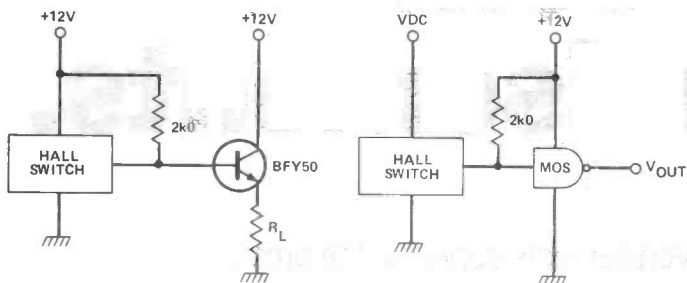


Fig. 14 (left) Driving CMOS from a Hall effect device.

Fig. 15 (right) A current amplifier for the Hall device.

supplies 80 mA to the triac gate to turn on the load current. It should be noted that the Hall device is connected to one side of the mains supply; this could be avoided by the use of an opto-coupling device between the Hall IC and the triac circuit.

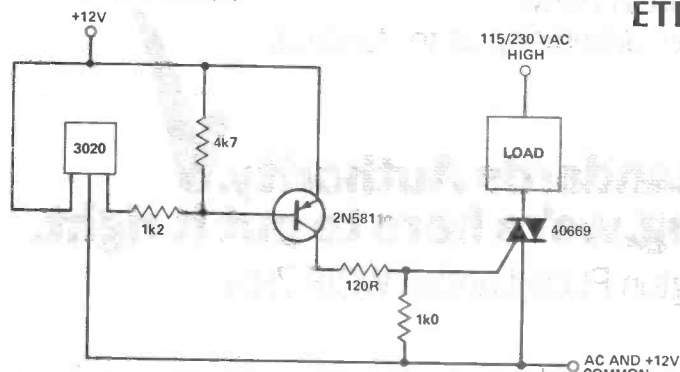


Fig. 16 AC power control using a Hall effect device.

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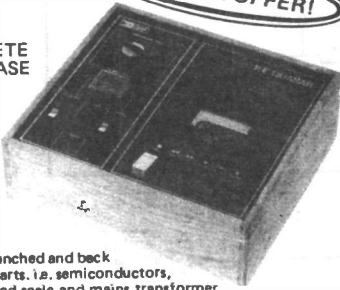
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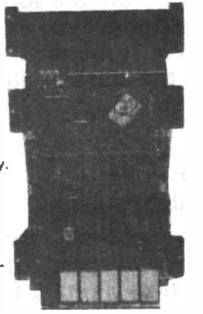


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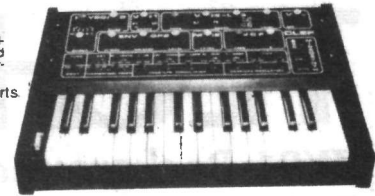


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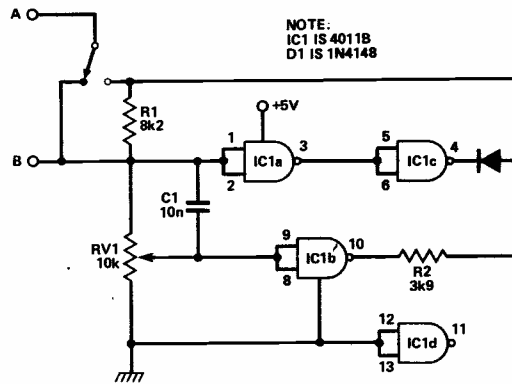
ZX81 Background Reverse

F.W. Picken,
Stoke on Trent

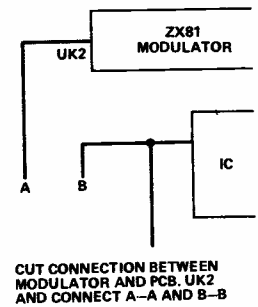
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IC1 IS 4011B
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CUT CONNECTION BETWEEN MODULATOR AND PCB. UK2 AND CONNECT A-A AND B-B

very low, and can be taken from the ZX81 supply line.

Action of the circuit is very simple: IC1a and c carry the negative-going sync pulse and nothing else, thus ensuring that this is not inverted. The remainder of the video signal is inverted by IC1b. The value of RV1 needs careful setting to obtain the best results.

Sophisticated 5V Supply

A.J.J. Gilchrist,
Paris

This supply will deliver up to 6A at +5V. It has an active overvoltage clamp on the output. Current is limited at around 6A and has foldback short-circuit protection.

Starting on the line side, S1 is a transient suppressor which helps protect the supply against high voltage transients (studies have shown spikes of 5kV to be common on the domestic mains!), and C1, R1 cut down on switch arcing when the supply is switched off. R2 is a bleed resistor which is useful if the supply is not permanently connected as it prevents the output staying high after the unit is switched off.

Q2 is the pass resistor, it is switched on by R3 when more than 100mA is being drawn from the supply. Q1 and Q3 provide the current limiting. When current through R7 is 6A the voltage on the base/emitter junction of Q3 is given by:

$$V = V_{out} - \frac{R5 * (V_{out} + R7 * 6)}{(R5 + R6)}$$

= 0.58V for the values shown.

Thus Q3 begins to conduct, switching Q1 on, which increases the base voltage of Q2, thus limiting the current.

This arrangement also provides foldback limiting. If we consider the output shorted, the voltage at the base of Q3 will be about 0.6V; thus the voltage at the collector of Q2 is approximately given by:

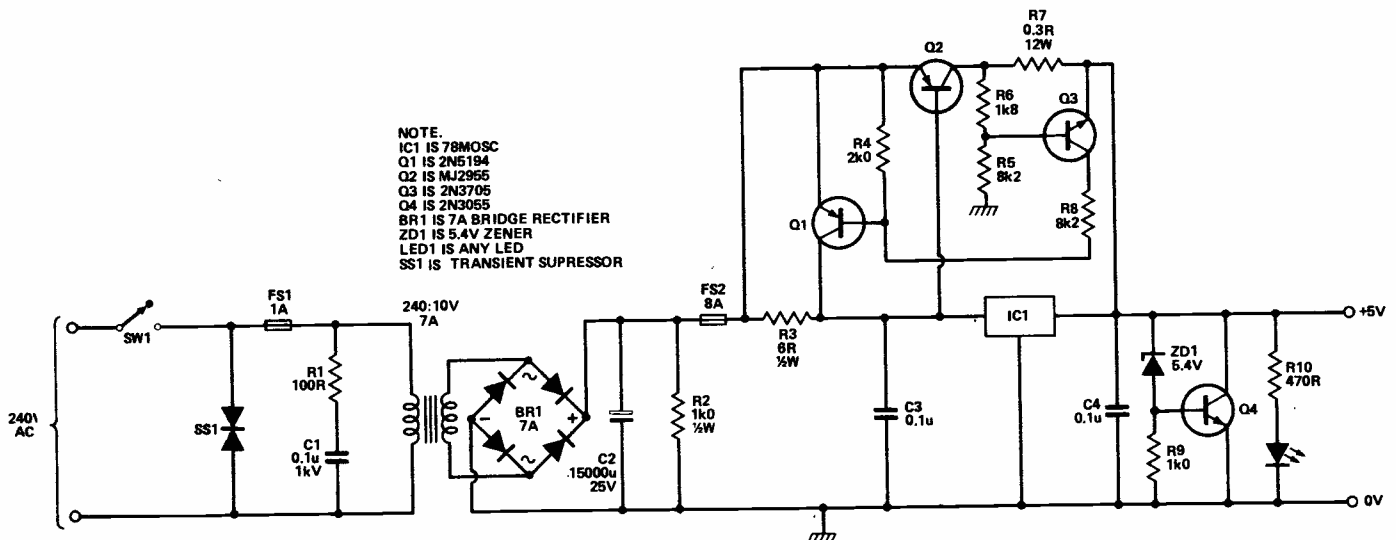
$$V = 0.6(R6 + R5)/R5$$

= 0.7V

thus the current supplied will be $i = 2.3A$. The foldback current may be changed by altering R5, R6 and R7. I have set this high to ensure the supply starts up under heavy loads.

Z1, R9 and Q4 provide an active overvoltage clamp. When the output voltage rises above the zener drop plus the switch-on voltage for Q4, Q4 conducts. This circuit is capable of sinking 15A indefinitely (with a proper heat sink) with a much higher peak current. This ensures that the fuse blows before the protection circuit. This active clamp is necessary because one of the most common types of failure in power supplies is the pass transistor failing with the collector/emitter junction shorted.

Note that Q1, Q2 and Q3 should all be mounted on heat sinks, and Q1 must be capable of passing the maximum output current of the 78M05 (500mA).



NOTE:
IC1 IS 78M05
Q1 IS 2N5194
Q2 IS MJ2955
Q3 IS 2N3705
Q4 IS 2N3055
BR1 IS 7A BRIDGE RECTIFIER
ZD1 IS 5.4V ZENER
LED1 IS ANY LED
SS1 IS TRANSIENT SUPPRESSOR

Ramped Pulse Generator For Stepper Motors

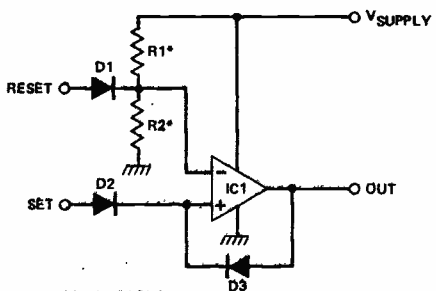
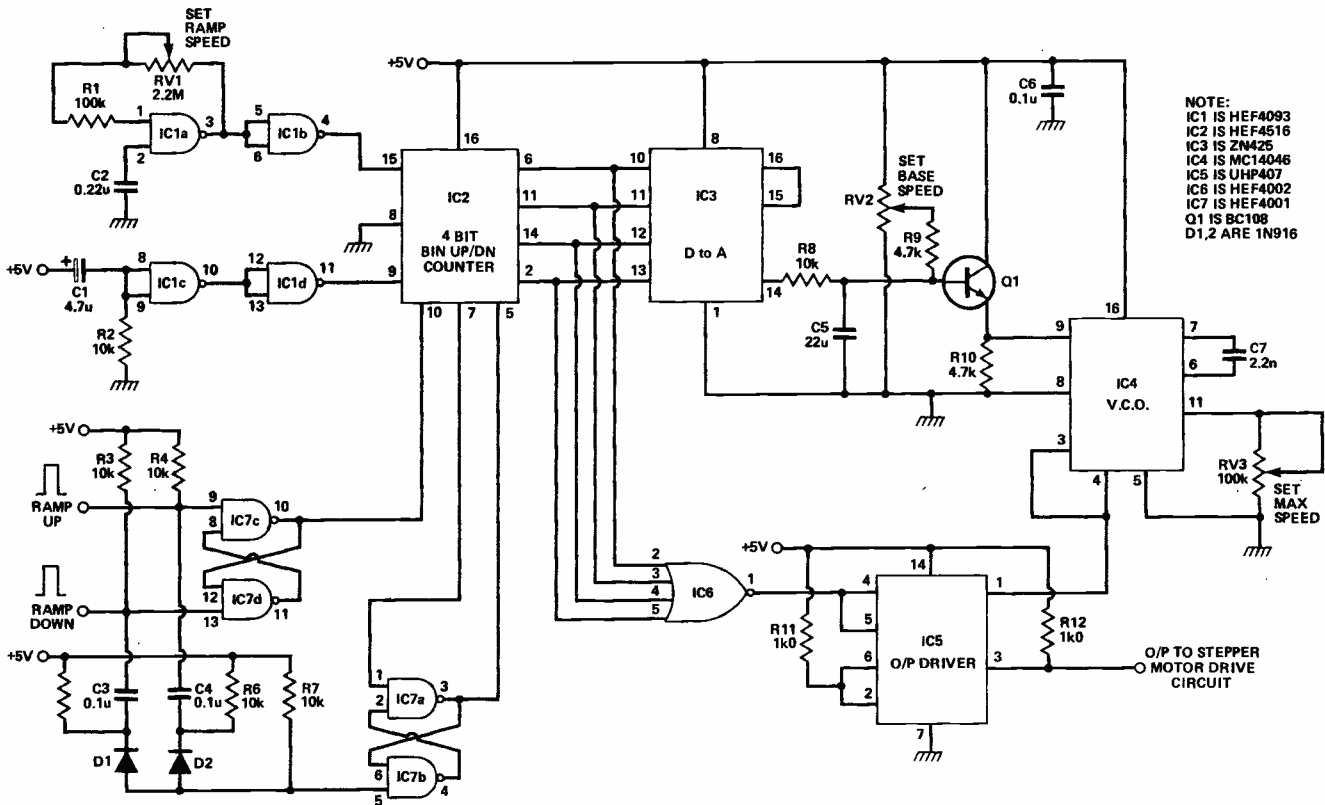
Clive Pantrey
Farnborough

A circuit that I needed for use with an experimental robot arm was one which would ramp up or down the pulse repetition rate of pulses delivered to a stepper motor drive circuit. With the motor stopped, the command 'ramp up' should start the motor at its base speed (ie the speed at which the motor will start and stop

under load, without loss of synchronisation), ramp up to maximum speed (the top speed available without loss of synchronisation) and run for as long as required. The command 'ramp down' should make the motor slow down to base speed and then stop.

The figure shows the circuit that was eventually produced. IC1a/b provide clock pulses (Ramp Speed) to a four-bit binary up/down counter, IC2. On receipt of a ramp up command IC7c/d sets count up and IC7a/b enables the counter. Unless a ramp down command is received the counter will reach its maximum count (Max Speed) and hold at this until the ramp down command is

received; this will set count down and enable the counter, which will count down to zero (Base Speed), and hold again until the next ramp up command. The counter output drives a D-to-A IC3 whose ramping output controls the VCO, IC4. The lower frequency of the VCO, (Base Speed) is set by the bias adjustment of Q1. Upper frequency (Max Speed) is set by the 100k pot at pin 11 of IC4. IC5 provides open collector drive for the output pulse train and also the on/off gate, controlled by IC6, when the counter is at zero. IC1c/d provides a set zero pulse to IC2, to ensure that the output, at pin 3 of IC5, is at base speed and off each time the generator is switched on.



TRUTH TABLE

SET	RESET	OUT
0	0	NC
0	1	0
1	0	1
1	1	1

NC = NO CHANGE

*SEE TEXT
NOTE: IC1 IS ANY OP-AMP
D1-3 ARE 1N4148

Analogue Set-Reset Latch

T. P. West, Lancaster

Although CMOS gates are commonly used to provide analogue amplifiers, the operational amplifier is often overlooked for use in digital applications. Often, a circuit design calls for a set-reset latch within an analogue digital circuitry to be included in the design. By the use of this cir-

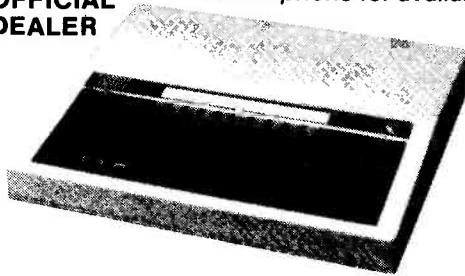
cuit, spare op-amps in a package may be utilised to provide the set-reset function. The op-amp used may be of any type with the low and high voltages at the output being only a function of the op-amp's internal output drive circuitry. The resistors R1 and R2 should be chosen so that $R2 = 2.4R1$ and $R2 < V_{supply}/0.05$. Although the circuit is shown for a single supply rail, it will work on a dual supply but produces a low of around the negative supply voltage. All changes in state occur on the low-to-high transition.

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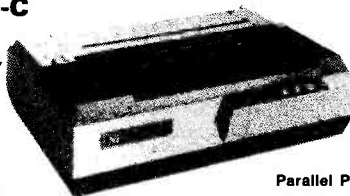
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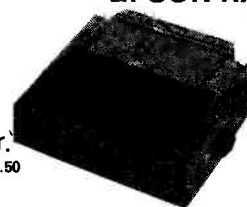
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Z80 CONTROLLER COMPUTER

The UK is now full of MARVINs just sitting there rusting, or worse, complaining about the pain in their diodes all down one side. Peter Grigson and David Harris tell us about the software needed to make them all spring into life.

The following is split into two sections; the first is a general introduction on machine code programming the Z80. Without writing an entire book, we can't really tell you all about it, so what we've aimed at doing is to give you a flavour of what's involved. There are quite a large number of books on the Z80 in particular, and on microprocessors in general, and we would suggest getting hold of one of these. Alternatively, elsewhere in this magazine is the start of a series on machine code programming.

The second section will move on to a brief description of the operating system for MARVIN that is pre-programmed into his EPROM.

Writing Z80 Programs

The nice thing about machine code programs is that they run the computer very fast and need a comparatively small amount of memory to achieve what can be a fairly sophisticated task. In general, machine code programming is much more appropriate to control functions because it enables you to tell the computer exactly how you want a particular task to be carried out.

Every silver lining has a cloud, and the cloud over using machine code is that a program consists of nothing but numbers, as has already been pointed out on page 35. However, common practice is to substitute mnemonics for the code when doing the writing, and convert the instructions into machine code using either a special program (an assembler) or a table and a great deal of hard work. Such is life . . . For the sake of clarity, we will use mnemonics for the remainder of this article.

Getting Into The Z80

Internally, the Z80 CPU has seven eight-bit registers, A (the accumulator) and B,C,D,E,H,L. There are three pseudo-16-bit registers, made by pairing B and C,

have different machine code equivalents, not just the same instruction with a different register address to be loaded.

Contents of registers can also be transferred. Any eight-bit register can be loaded with the contents of

0000-07FF	MARVIN operating system
0800-0FFF	User program. The operating system hands over to the routine at 0800 after reset and so this should contain the first instruction of your program.
8000-83FF	RAM. The first 32 bytes from 8000 to 801F are used by the operating system and the stack extends down from 83FF. The rest is freely available.

Fig. 1 Memory map of MARVIN.

D and E and H and L. There are also the two 16-bit index registers, IX and IY. Each register can only be used with certain commands and it will pay you to make yourself familiar with the rules. All registers can be loaded directly with a number, and the mnemonic for this will be **LD (reg), (number)**, where (reg) is the register to be loaded and (number) is the binary (or hexadecimal) number to be stored; for example, **LD A, 23** stores 23 (hex) in the register A; **LD HL, 1234** stores 1234 in the HL pair as a 16-bit number.

Remember that the above is written in mnemonic code, and that unless you have another micro to do the work for you, you will have to translate this into machine code before MARVIN will be able to understand what it means. Also worth noting is that the two instructions **LD A** and **LD HL** will

another. For example, **LD B,D** copies the contents of D into B. Sixteen-bit registers can only be copied by two operations on their component registers; to transfer HL to BC **LD B,H** then **LD C,L**.

You will very often need to operate with more numbers than the available registers can hold and so numbers must be transferred between RAM and the CPU registers. The accumulator or any 16-bit pair can be stored at a specific address. For example, the instruction **LD (8234), A** stores contents of A in RAM at address 8234. The instruction **LD (8234), HL** involves storing 16 bits so L goes into memory at 8234 and H at 8235.

Any of the other eight-bit registers can be stored at the location pointed to by the contents of HL: **LD HL,8349** then **LD (HL),E** puts the contents of E at 8349.

OUT (n),A	Output accumulator contents to port n. $0 \leq n < 10$.
OUT (C),r	r is any of A,B,C,D,E,H,L. The contents of C register define the port. $0 \leq C < 10$.
IN A,(n)	Load accumulator with data on port n. $0 \leq n < 10$.
IN r,(C)	r is any of A,B,C,D,E,H,L. The contents of C register define the port. $0 \leq C < 10$. Load register r with the data on port C.

Fig. 2 The Z80's Input/Output instructions. Note that it is recommended that output is channeled through the operating system if you wish to operate on individual bits.

Another useful source of temporary storage is the stack. This is a 'pile' of 16-bit numbers in RAM onto which more can be added by the instruction **PUSH: PUSH HL** puts the contents of HL onto the top of the stack. The number on top can be retrieved by the instruction **POP**, eg **POP DE** removes the number from the top of the stack and puts it in DE. A special CPU register, SP, is used to point to the stack and is updated by each **PUSH** and **POP**.

Only simple arithmetic and logic instructions can be carried out. Any eight-bit number or register can be added to, subtracted from, ANDed with, XORed with, ORed with the accumulator and the result is stored in the accumulator, eg **ADD A,27 ; SUB A,C ; AND A,E ; OR A,H ; XOR A,255**.

As far as 16-bit registers are concerned, any of them can be added to or subtracted from HL,IX,IY with the result being stored in HL,IX,IY. For example, **ADD HL,DE** adds the contents of HL to the contents of DE and stores the result in HL.

The order in which the code is carried out can be controlled by use of **JP** (equivalent to the BASIC GOTO) and **CALL** (equivalent to GOSUB). When using an assembler, various points in the code can be identified with labels by writing the label at the beginning of the line with no preceding gap and following it by a colon. This tells the assembler to assign the value of the program counter at that point to the label so that when the label is used in conjunction with **JP** instructions, the appropriate 16-bit address is assembled. Labels can also be used in conjunction with other instructions that require 16-bit addresses, for example, **LD A, (LABEL)**. The instruction **JR** is used for local jumps within 128 bytes and assembles to a relative displacement instruction. **RET** is used to return from a CALLED routine.

Decision making in machine code is carried out by the use of flags. These are set according to the result of each logical or arithmetic operation. The most useful flag is called Z (the zero flag) and is set if the result of an operation is zero. It is also set if A equals B in compare instructions, eg **CP A,B**. There are several other flags indicating other conditions and you should refer to a book to find out what these are, and when they are set.

The flags are used in conjunction with **JP, CALL, JR, RET**

which can be made conditional on the state of a particular flag. For example, after executing the instructions **CP A,3** then **JP Z,FINISH** the processor will only have gone to FINISH if A was equal to 3.

The most useful instructions on MARVIN however are those involved with I/O. There are several to choose from, and Fig. 2 shows them. Note that it is best to avoid using the output instructions directly but to use the operating system output routines which will also correctly set the port masks (see below).

In order to follow the example program in Fig. 3 you will need to know one or two more things. An instruction called **LDIR** is used which shifts blocks of memory around. It takes a number of bytes from memory starting at the location defined by the contents of HL and copies them to memory commencing at the location defined by the contents of DE; the number of bytes it copies depends on the contents of BC.

DEFB is not a true Z80 instruction but one to the assembler. It is followed by a series

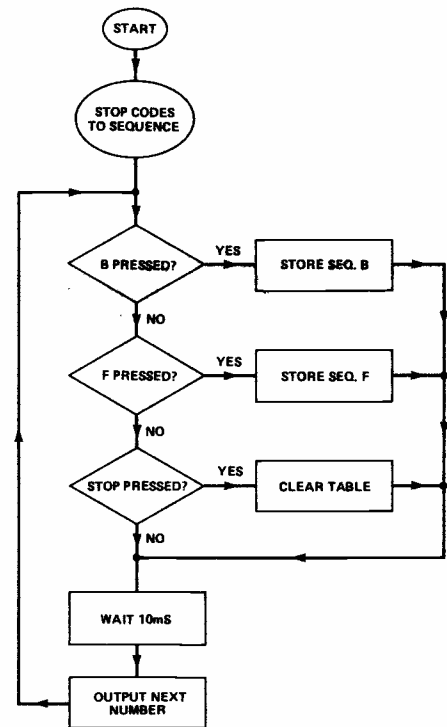


Fig. 3 An example program to drive a stepper motor. The two operating system routines used are 5, which reads the keyboard and returns a number at location 801BH, and 7, which outputs successive numbers in a sequence each time it is called. IX is the beginning of the table; IX + 0 is the table; IX + 1 is the position in the table of the first one to be output; IX + 2 is the first byte in the sequence, etc.

	ORG	800H	Set start of prog. to 800H in memory
	JR	STOP	Stop motor at beginning
NKEY:	DEFB	0F7,4	Instruct processor to execute op. sys. routine 4
	LD	A,(801BH)	Number of key returned to 801B — transfer to A
	CP	A,0BH	Compare content of A with 0BH
	JR	Z,BAK	If comparison true go to routine at BAK
	CP	A,0F	Compare with 0FH
	JR	Z,FWD	If true go to FWD
	CP	A,00	Compare with 00H
	JR	Z,STOP	If true, go to STOP
NEXT:	LD	BC,990	Start of 10 mS pause; for 10 mS put 990 in BC
	DEFB	0F7,5	Call op. sys. routine which waits 100 + 10BC ms
	LD	IX,8020H	Set pointer to beginning of table
	LD	C,1	Sequence to be output from port 1
	DEFB	0F7,7	Call op. sys. routine to output next byte
	JR	NKEY	Return to start
BAK:	LD	DE,8020H	Start of sequence table in RAM
	LD	HL,BTAB	Point HL to table in ROM
	LD	BC,6	6 bytes to be copied
	LDIR		Copy the backward sequence to RAM
	JR	NEXT	Goto NEXT
FWD:	LD	DE,8020H	Start of sequence in RAM
	LD	HL,FTAB	Point HL to table in ROM
	LD	BC,6	6 bytes to be copied
	LDIR		Copy the forward sequence to RAM
	JR	NEXT	Go to next
STOP:	LD	HL,8020H	Fill RAM sequence with 0s
	LD	DE,8021H	
	LD	BC,6	
	LD	(HL)0	
	LDIR		
	JR	NEXT	
FTAB:	DEFB	4,0,10,6,9,2	
BTAB:	DEFB	4,0,2,9,6,10	
	END		

of eight-bit numbers which the assembler places directly in memory. This can be used in MARVIN programs to assemble the special instructions for calling the operating system routines which consist of two bytes: F7 (hex) followed by the number of the routine, eg **DEFB F7,03**.

An example of a typical MARVIN program is shown in Fig. 3. This program turns a motor forwards if key F is pressed, backwards if key B is pressed and stops it if key O is pressed. It uses several of the operating system routines described below.

A stepper motor is connected to output port 1 using power transistors so that to turn it forwards the sequence

10, 6, 9, 2

must be sent, one number every 10 mS. In order to turn it backwards the reverse sequence must be sent. The motor stops if nothing is sent.

The Operating System

The operating system has two tasks: it controls the system, and this involves such chores as dealing with start-up initialisation, dealing with interrupts, communicating with other systems, etc; the other task is to provide various useful routines for controlling MARVIN's peripherals.

When power is applied or the reset switch is pressed, the operating system first clears all ports and sets the interrupt vectors to the beginning of the user program. It also sets the stack pointer to the top of RAM (83FF). It then tests to see if the user EPROM socket contains a RAM IC or an EPROM. If it detects the latter then control is transferred to the program in the EPROM at 0800H. If it detects a RAM then it enters a routine which can receive data from a microcomputer and place it in the RAM. In order to be suitable for connection your micro must have an eight-bit-user output port and either a separate user input port or a single handshake input line; Z80 PIO and, 6522 VIA are suitable. This facility is extremely useful for testing prototype programmes without having to blow an EPROM each time.

Interrupts

In order to define MARVIN's response to interrupts, the user program should place eight vectors in RAM to define the start of routines to deal with interrupts on

each channel. Interrupts can then be enabled by an EI instruction and the appropriate routine will be called (with all registers and flags preserved) on receipt of an interrupt.

User-available routines include the following:

Output Port Control Since the Z80 only provides for alteration of all eight output bits of a port at once, if the user wishes to change a bit or bits without affecting any others a note must be kept of how each bit has previously been set. The operating systems deals with this by storing masks of each output port in RAM and, if the user ensures ports are written to via the operating system routines, then the masks will be constantly updated. Routines are provided for the setting/resetting of individual bits as well as whole ports.

Outputting a sequence Certain devices such as stepper motors require a sequence of bytes to be output in order to operate. The operating system provides a routine which will output consecutive bytes of a sequence each time it is called.

Timing If a real time clock is not available then the operating system is capable of generating pauses of between 100 μ S and 0.65 secs to the nearest 10 μ S and also between 0.1 and 25 seconds to the nearest 0.1 seconds.

Peripheral Control Routines are provided to : 1. read a keyboard of up to 64 keys consisting of simple switches connected between an input and output port ; 2. display alphanumeric data on a 7 segment display connected to an output port ;

BUYLINES

The following will be available from ARK Electronics, 3 Barnhill, Pinner, Middlesex, HA5 2SY (please note this change of address).

EPROM containing the monitor program, 4 MHz clock, £6.00; 3 MHz (or lower) clock £4.00;

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I/O Board PCB £1.50;

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Because the remaining components for the I/O and interrupt boards are very easily obtainable, ARK have not judged it worthwhile making kits available from them.

3. detect the last code sent by the MARVIN remote control ; 4. send allophones to the MARVIN speech board ; 5. output a frequency burst of between 40 Hz and 10 kHz.

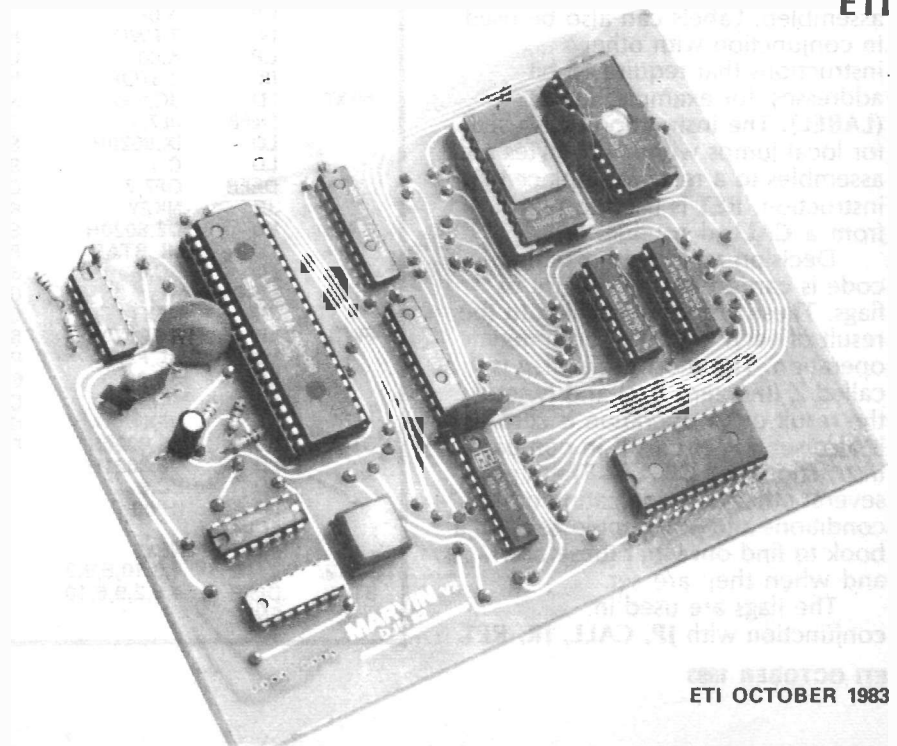
The other peripherals available (D/A, A/D, light dimmer) are accessed simply by writing to or reading the relevant port.

All the routines are accessed by 2 byte routines of the form : F7, nn where nn is the number of the routine in question.

Full details of exactly how to use all the routines are supplied with the operating system EPROM.

Readers who have been puzzled by the naming of MARVIN, or who have found little sense in the humour adopted in this project series (and throughout much of the rest of the magazine) are referred to 'The Hitch Hiker's Guide To The Galaxy' et seq., copies of which should be on sale from all leading component suppliers in the Alpha Centauri district.

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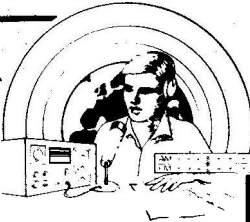
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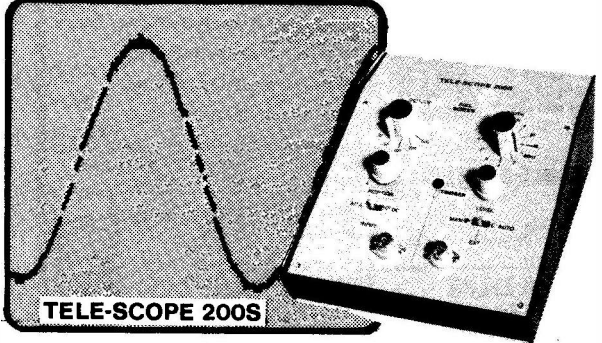
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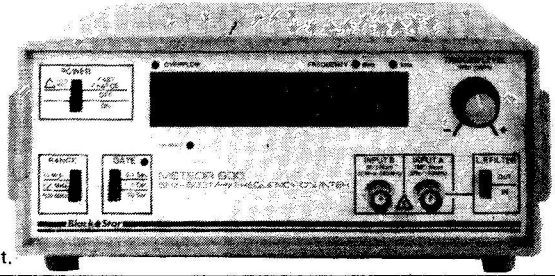
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DON'T SNEEZE!

There are places on this earth where a sneeze can be very expensive indeed — and where spotlessness will be only just clean enough. Stephen McClelland elucidates.

An anecdote, popular in the silicon chip industry, relates how, in the recent past, a certain chip company found its production going haywire. The devices it made suddenly became duds because of contamination of the otherwise ultra-clean process areas, but no one knew where the trouble lay. At length, after a massive investigation during which the factory was almost taken apart, the Company discovered to its considerable embarrassment that the person responsible for cleaning the operators' overalls had changed the washing powder. Contaminant had migrated from the freshly washed overalls into the process line, and so to the chips themselves.

Although an extreme case, the tale serves to illustrate how careful chip manufacturers have to be to ensure their production environment is flawless and the paranoia that descends when it isn't. Setting up these facilities, the cleanest places on the planet, is prodigiously expensive — upwards of £300 per square foot.

Such facilities are now becoming necessary because the complexity demanded of silicon chips takes them into the VLSI (Very Large Scale Integration) domain. These chips have features, therefore, which are unprecedentedly small — about 1 micro metre (or 0.04 thousandths of an inch) in most cases. As a result even the most minute particles of dust are capable of settling on the silicon circuit during its sensitive process stages and causing havoc.

Indeed there is evidence to suggest that even the maximum pure air conditions presently achieved might not be clean enough for future chips under some conditions. But even to get to this state (dubbed Class 100 by the US because it contains a bare 100 particles per cubic foot in the critical size range) requires enormous efforts. Class 100 requires a clean-up factor of more than 100,000 times on atmospheric air (containing about 10 million particles per cubic foot).

Such clean rooms are generally maintained at positive atmospheric pressure (to prevent outside air-borne dirt from being driven in) with a wide host of complicated access doors and passages. In the cleanest facilities, floors and ceilings are both perforated to allow the filtered air to be pumped through them downwards in parallel vertical lines. The laminar flow creates a minimum of turbulence which might otherwise re-distribute the particles already present.

In addition, these areas need a formidable battery of support services. The water used to wash the silicon chips themselves must approach absolute purity — and the inadequacy of most conventional distillation techniques means that the water is usually the most expensive chemical the plant has to purchase. The gases used to pro-

cess the silicon can also present a hazard — either because they are toxic, or flammable, or both. This is one reason why even small semi-conductor houses in California — especially those located near the San Andreas fault — maintain elaborate fire fighting teams.

But getting a clean room 'clean' is only half the story — keeping it so is rather more difficult and research throws an interesting light on the most significant source of dirt — people themselves.

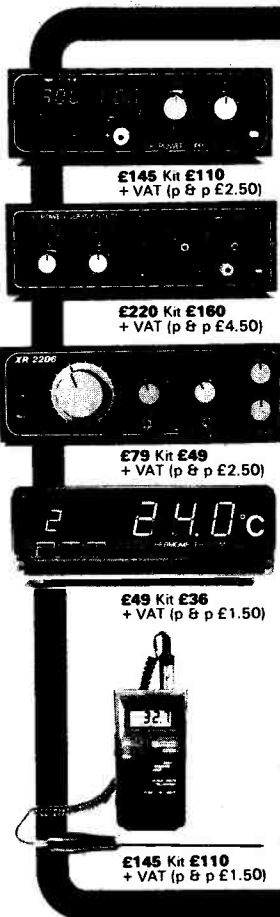
A NASA study conducted by James Useller turned up some surprising conclusions. He found that one of the most destructive things you can do in a Class 100 environment is simply to write on an ordinary sheet of paper with an ordinary ballpoint pen — this alone can generate particles up to 20 micrometres in size. Stamping on the floor (whether due to a fit of pique or just through lack of exercise the study neglects to say) is almost as bad. Even workers properly clad — in astronaut-like boots, trousers, smocks and hats — can generate or redistribute particulate matter by merely moving about irregularly.

Normal breath, in fact, should produce no increase of atmospheric dirt but the best advice one can give to smokers is that they shouldn't breathe at all — NASA detected significant numbers of particles 20 minutes after the subjects had finished smoking. Personal hygiene is encouraged. Anti-social habits like scratching yourself (which releases dead skin cells) are not. Sneezing is singled out as being a big offender — it can push the particle count up by as much as 20 times. Even an overtly social act like pulling out a handkerchief to suppress it can spray the same amount of contamination into the air.

An equally insidious source of contamination is the purely chemical kind as the story at the beginning shows. Sodium ions (whether from sweat, chemicals or soap powder) can be lethal to many types of devices particularly the MOS (Metal-Oxide-Silicon) varieties. Some engineers have even suggested that MOS chip factories should never be built near sea coasts because of possible salt contamination.

Silicon chip companies are aware that the contamination risk will eventually become so strong even with conventional clean rooms that chip making will have to be taken — quite literally — out of human hands. IBM has already gone some way towards making an automatic production line that is enclosed completely in its own clean environment. Other companies are deterred by the enormous amount of investment required to overcome the tricky problems of silicon water handling. Ironically, micro-electronics, of all industries, might be very difficult to automate fully.

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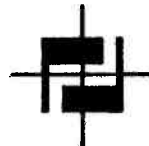
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ZX80 TAPE MOD

When we published the article on modifying the ZX81 tape output to get reliable cassette operation, a plaintive cry went up: "What about my ZX80?". Ian Ridout to the rescue.

The predecessor to the ZX81, the ZX80, is still a popular machine especially if fitted with the replacement Sinclair ROM. However, like the ZX81, it suffers from unreliable SAVEing onto cassette tape and LOADing from tape.

A previous article in this magazine (February 1983) gave details of how to improve the reliability of SAVEing programs onto tape for the ZX81. This article gives details of the simple modifications required to allow the ZX80 to enjoy the same SAVEing reliability onto tape. The modification costs less than £1 to do and takes only a few hours.

If you have experienced difficulty with SAVEing programs onto tape with your ZX80, observe the following recommendations before implementing this modification:

- Make sure that the EAR socket on the ZX80 is connected to the EARPHONE socket on the cassette and likewise that the MIC socket goes to either the MICROPHONE or AUX socket on the cassette (we don't want to insult your intelligence, but you'd be surprised how often . . .);
- Remove all traces of the brown magnetic tape material from the heads in the cassette player by using cotton buds dampened (not dripping wet) in white or surgical spirit;
- Use computer tapes or the higher quality audio tapes;
- When SAVEing and LOADing, keep the cassette player as far as possible from the television because the television line-scan and frame-scan signals will be picked up by the cassette player;
- On playing back from tape (ie LOADing) keep the volume as high as possible without the television picture breaking up. If the picture begins to break up, this indicates that the input level is overloading the computer circuitry and the playback volume on the cassette should be slightly reduced. Having

found the correct level, rewind the tape to the beginning and try to LOAD the program into the ZX80 again.

If these recommendations still fail to give you reliable SAVEing and LOADing then this article could well help you. The problem is probably due to the low signal level coming out of the MIC socket when SAVEing.

The tape system in the ZX80 is very similar to that in the ZX81, the major difference being the component numbering! As before, one IC output is for both the TV modulator and the tape out socket, via a simple filter with a peak at 3.4 kHz and a roll-off of 6dB per octave on either side of the peak. Like the ZX81 filter, the loss is -66dB at the pass frequencies, which leads to only 2 mV of signal at the tape output.

Circuitry Modification

The reasons behind choosing a FET for this modification were discussed in the previous article, so I won't repeat them here. The new circuit differs slightly from that used on the ZX81 to take account of the different PCB layout.

The DC biasing conditions of the FET are such that the source voltage should be between 1 volt and 3.5 volts and this is achieved by making the source resistance about 4k and the drain current about 0.5 milliamps.

C13 (47nF) and R34 (1k Ω) are retained in their original positions but C14 (47pF) is moved and R35 has to be changed in value only if a manual record-level tape recorder is used. For use with an automatic record-level machine R35 (1M Ω) is retained but one end has to be desoldered from the printed circuit

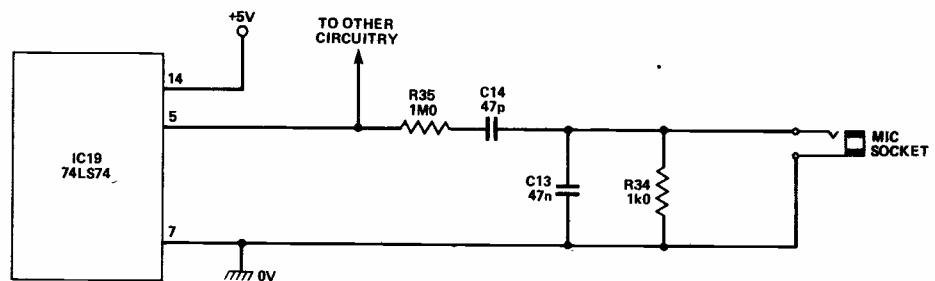
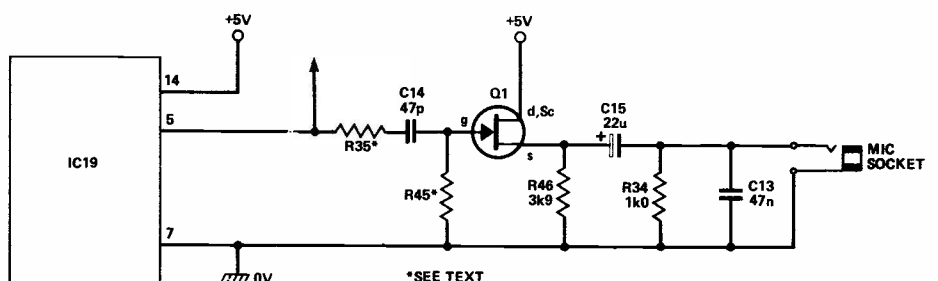
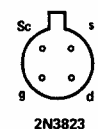
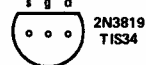


Fig. 1 Original tape-recording circuit.



NOTE
Q1 IS 2N3819, 2N3823, TIS34, ETC



FET CONNECTIONS
(VIEWED FROM PINS)
s = SOURCE
g = GATE
d = DRAIN
Sc = SCREEN

Fig. 2 New circuit; note the difference between this and the circuit published in February.

board. See Table 1 for the values of both R35 and the gate bias resistor that I have called R45.

Most machines used for storing programs on tape are of the automatic record level type. If, with the resistor values shown for R45 (39K), the recording sounds distorted on playback through the cassette loudspeaker, R45 should be reduced in value to 6k2 or 8k2. The sum of the values of R35 and R45 should be within the range 900k to 1M1 to preserve the filter characteristics.

Doing The Modifications

Remove the five white and two black plastic studs holding the bottom half of the case to the top half and to the keyboard. These are removed by first pushing the centre plastic pins through the outer part of the studs and then pulling the complete stud out of the case.

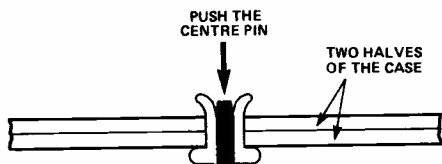


Fig. 3 How to get inside the ZX80.

Now remove the three plastic clips holding the PCB to the back half of the case by carefully squeezing them and pushing them through the PCB.

For use with manual record-level machines, remove R35 (1M0) and insert one end of the new R35 (820k) into the left hand end of the position vacated by the old R35. For use with automatic record-level machines, desolder only the right-hand end of R35 (1M0). The rest of these instructions apply irrespective

of the type of cassette machine to be used.

Carefully remove C14 (47pF) and resolder it in series with R35 so that its right-hand end is soldered into the PCB hole from which the right-hand end of R35 was removed. The left-hand end of C14 should be soldered to the free end of R35 and trimmed off as short as possible.

Solder R45 (See Table 1 for value) in the position shown in Fig. 5. Solder the negative end of C15 (22 μ F, 16 volt) as shown keeping the lead short and leave its positive end unconnected for the moment.

If the metal cased 2N3823 is used, solder the drain and screen leads together and then cut one of them off. Solder the drain lead to the +5 volt PCB line shown. Connect the FET gate lead to the junction between C14 and R45. One end of a 3k9 resistor (R46) is soldered to the 0V lead shown and its other end connected to the positive end of C15 and the source of the FET. It is important to check the connections to the FET and to make sure they are not touching each other.

Put the PCB and the back of the case together again, securing them as before, with the three push-through plastic studs inserted from the back.

At this point test the modification by plugging in the television (tuned to channel 36) and the computer power. When the normal computer television picture appears, plug in the two cassette leads to the computer and the cassette, write a two or three line program and SAVE it. LOAD it into your ZX80 observing the suggestions earlier in this article.

Fault Finding

The FET gate should be at 0 volts, the drain should be at +5 volts and the source at about 2 volts (1V to 3.5V). If not, check the connections and the layout.

SAVE a short program on to the cassette and listen to it through the cassette loudspeaker. If it sounds distorted, reduce the value of R45 as mentioned earlier.

See the February '83 issue of this magazine for a fuller fault finding guide.

PARTS LIST

This is a list of the new components required:

RESISTORS ($\frac{1}{4}$ w 5%)

R35 see Table 1.
R45 see Table 1.
R46 3k9

CAPACITOR

C15 22 μ F, 16 v axial electrolytic.

SEMICONDUCTOR

Q1 2N3819, TIS34 or 2N3823

TABLE 1

	R35	R45
Manual Recorders	820K	220K
Auto-level Recorders	1M	39K*

*See text

Table 1 Values of R35 and R45

BUYLINES

Nothing, but nothing should cause you any problems here. You don't even need a PCB!

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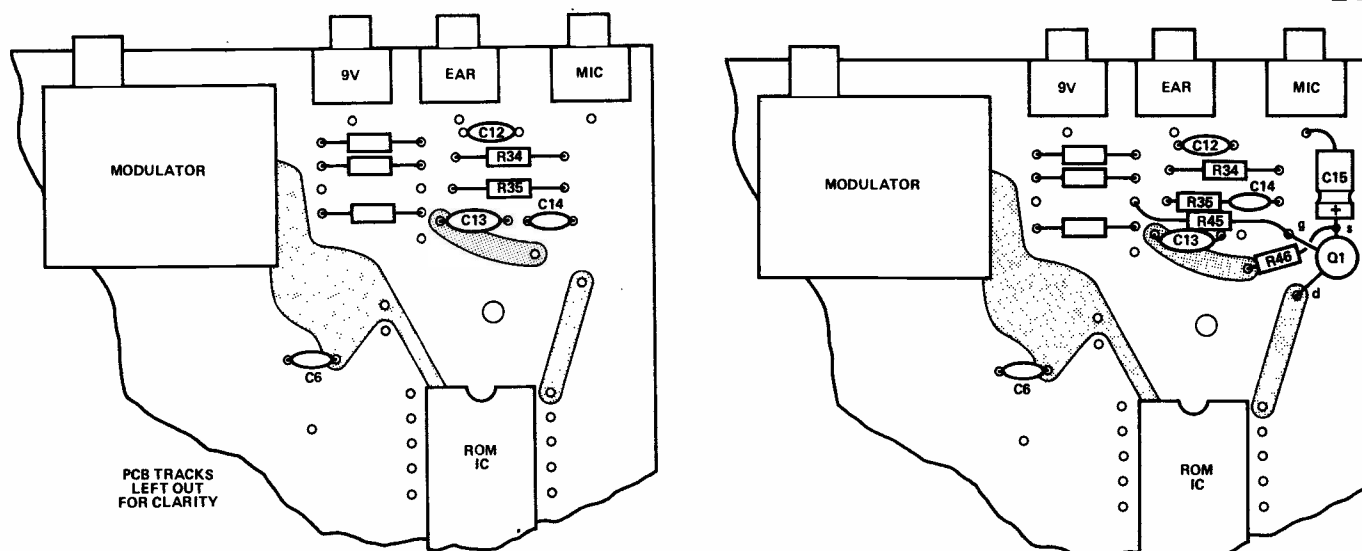


Fig. 4 Before (left) and after (right) the modification.

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THANDAR TF040 10 Hz to 40 MHz
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CAR ALARM

Alarmed at the prospect of your car taking a walk? Ian Forster shows you how to protect it, or your house, a little better. Development by Phil Walker.

This multi-purpose alarm unit can be used in a car, or, if desired, in the home with the addition of a 12V power supply. The basic version provides for a fairly simple alarm that can be

triggered either by a negative-going or a positive-going pulse.

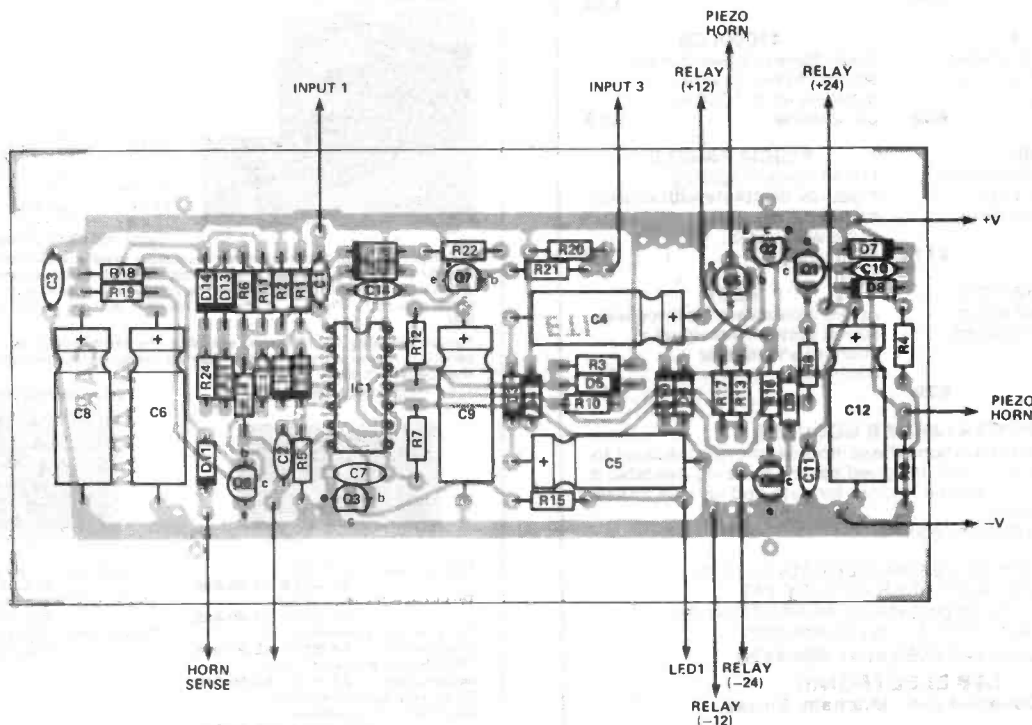
If the source of the alarm is removed, the unit will reset itself, as is required by law (and one's neighbours!). There is a delay of

approximately 16 seconds before sounding a piezo alarm, which could be used as a warning, and the main car horn will be sounded after about 24 seconds.

There seems to be a good

RESISTORS (all $\frac{1}{4}$ W 5%) R1, 2, 5 56K R3 560K R4, 8 10k R6 470k R7 100k R9 1k0 R10, 12 270k R11 820k R13 18k R14 33k R15 680R <i>R16, 17, 18, 19, 20,</i> <i>21</i> 18k R22 100k R23, 24 2k7	CAPACITORS C1, 2, 3 100n ceramic or polyester	C4, 6, 8, 9 33 μ 16V axial electrolytic C5 100 μ 16V axial electrolytic C7 100n or 1 μ unpolarised (see text) C10, 11 100n ceramic or polyester, 24V min C12 100 μ axial electrolytic, 24 min C13, 14 100n ceramic or polyester	SEMICONDUCTORS IC1 4093BE Q1, 2 BC212 Q3 BC182, TIP31, BD131	Q4, 6 BC182, TIP31, BD131 Q5, 7 BC212 D1-6 1N4148 D7-10, 12-15 1N4148 D11 OA91 LED1 single LED	MISCELLANEOUS RLA1 12V (or 24V, see text) relay, NO contacts	<i>Piezo tweeter or piezo horn (see text), normally open push-button switch, PCB, wire, etc.</i>	NOTE: Items in italics are for the extension options.
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Fig. 1 Circuit diagram: the basic circuit is shown in black and the optional sections in blue.



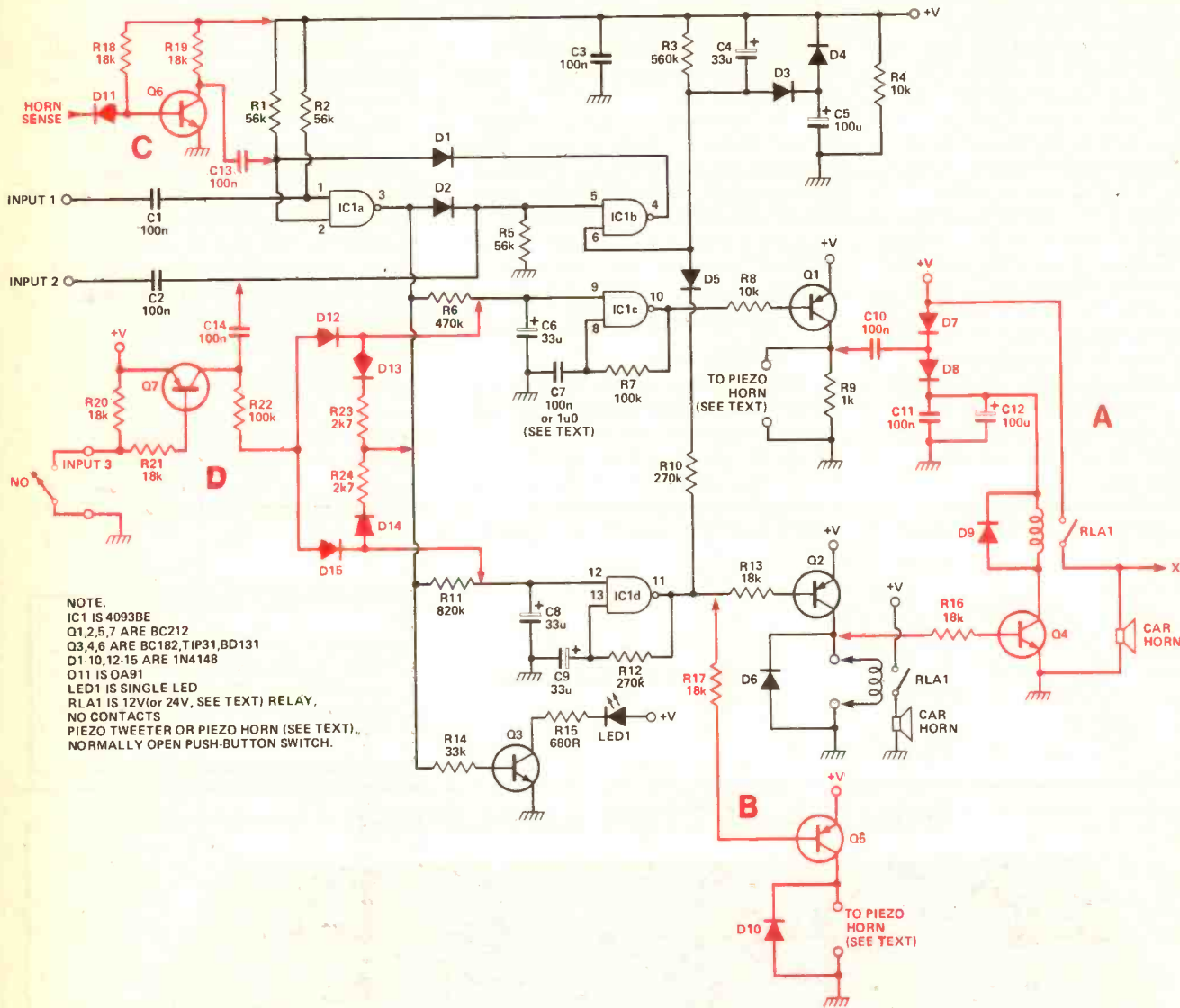


Fig. 2 Component overlay for the PCB.

HOW IT WORKS

The basic circuit is that of a latch, built round IC1a and IC1b, and two drivers for the audible alarms, built round IC1c and Q1, and IC1d and Q2. However, it's not quite that simple . . .

On switch-on, C4 and C5 share current via D3 pulling one of the inputs of IC1b low, which disables the latch until C5 is charged and C4 discharged via R3: this prevents the circuit being immediately activated at switch-on. If any troubles develop in this respect, it may well be because the values of C4 and C5 are not correct, electrolytics having very wide tolerances. On switch-off, C5 is discharged via D4 and R4.

A negative-going pulse at INPUT1 or a positive-going pulse at INPUT2 will trigger the latch, causing the output of IC1a to go high. C6 charges via R6 and C8 via R11, and LED1 is lit via R14, Q3, R15.

When C6 reaches a sufficient voltage, IC1c will start to oscillate at a frequency determined by C7/R7. For use with a piezo tweeter as a horn, C7 should be

100n which will give an oscillation frequency of 3 kHz. This can be fed to a piezo tweeter via R8 and Q1, to make a very unpleasant sound in a would-be thief's ear! R9 is needed to discharge the tweeter because there is no DC path through these beasts, and it would otherwise just sit at around +V.

Similarly, when C8 reaches a sufficient voltage, IC1d will begin to oscillate at about 1.5Hz, and this will turn the car horn on and off via R13, Q2 and RLA1. D6 protects Q2 against back-EMF from the coil of RLA1. The output of IC1d is also used to pump down C4 (via D5 and R10) so that the latch will be reset after a period, but will retrigger almost immediately if the input conditions persist.

Circuit A allows a 24V relay to be used instead of a 12V one. C7 should be changed to 1u0 to lower the oscillation frequency of IC1d to 300 Hz. C10, D7, D8 and C11 and C12 form a fairly straightforward voltage doubler to provide the 24V necessary. Q4 is needed to

invert the output from Q2 and pull down rather than up.

Note that in either case, you connect the relay contacts to suit your car — we have shown how to connect them when your car switches on the positive line to sound the horn, but some (eg Minis) switch the negative side, the positive side of the horn being permanently connected to the 12V supply.

Circuit B can be used to substitute a self-oscillating piezo-horn for the piezo tweeter — this could be useful if you use a 24V relay with circuit A.

Circuit C passes a small current through the car horn; if the connection to the horn is interrupted, the alarm will be triggered.

Circuit D is the fast-acting circuit; when the normally open external contact is closed, Q7 is turned on, and this charges C6 and C8 via R22, and D12 and D15, as well as latching the alarm in the usual way through C14. D13 and 14 and R23 and 24 are necessary to reset this circuit.

PROJECT : Car Alarm

supply of 24V relays around at bargain prices, mainly because they're a lot less convenient for most circuits than 12V types. For this reason, the circuit can be adapted to drive a 24V type, as shown in circuit A (shown in blue on the main circuit diagram). This circuit has the disadvantage that you cannot use the piezo tweeter, and circuit B shows how to connect a self-oscillating piezo horn instead.

A common thieves' tactic is to cut the lead to the horn before attempting to break into the car itself. This can be foiled by circuit C, which will trigger the alarm if this is done — note that once the horn lead is cut, there is no way that it can be used, so just the warning siren will go off.

Finally, there may be some items that you may want to protect

with a fast-acting alarm, and circuit D provides for this: all that is needed is a set of normally open contacts that can be arranged so as to close.

The method you use to trigger the car depends on how your car is wired up. Most cars have courtesy light switches mounted in the door pillars, and these could be used. Another possibility is to use mercury switches to detect any motion or disturbance.

Construction And Setting Up

Construction of the PCB should present no problems, provided the usual precautions are followed. Things being what they are, it would also be best to make up your mind which options you want before building, though at a pinch

bits and bobs can be added later.

As mentioned in How It Works, should there be any problems with the circuit arming at switch-on, then C4 and C5 are the most likely culprits.

If you use circuit A with a particularly current-thirsty relay, then it may be possible that R9, C10 and C12 will need revising (R9 should be reduced in value, C10 and C12 should be increased); Q1 may possibly need upgrading to a high gain medium power type and/or heatsinking, as might Q4.

Otherwise, your biggest headache will be the triggering switches. Unfortunately, we can't advise you generally on this because everybody's car or home will vary; but let us say that careful thought put into this aspect will not be wasted.

ETI

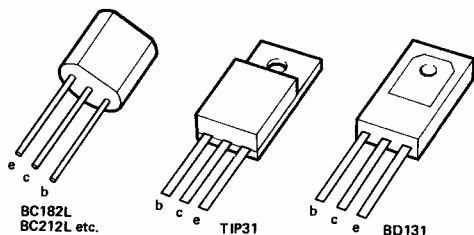
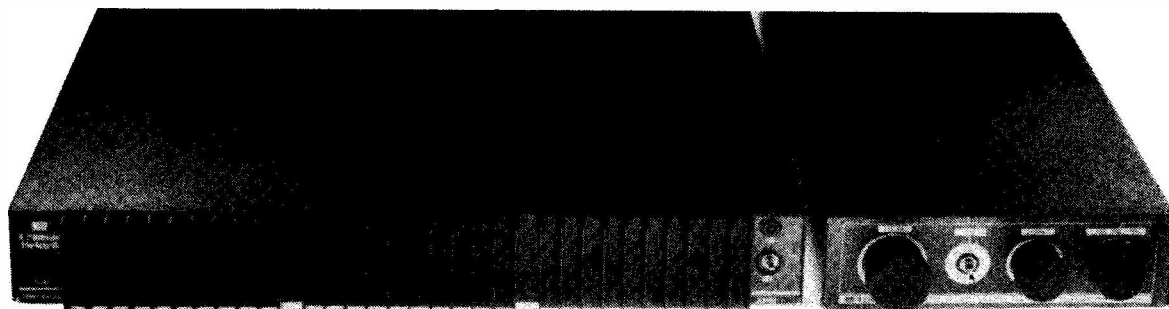


Fig. 3 Transistor pinouts.

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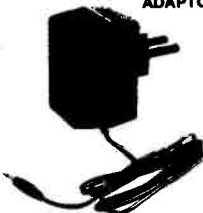
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HOW PERMANENT IS PERMANENT?

With the silicon revolution in full swing, everybody from private individuals to the governments of nations are dependent on the use of magnetism for the storage of information. Is our faith in the technique misplaced? Vivian Capel investigates.

Some years ago it was reported that the BBC were considering whether or not to dispose of their vast library of sound recordings on disc after dubbing them on to tape. The decision was against, because although there would be considerable savings in space, magnetic recordings were deemed to be too ephemeral to trust as the medium for preserving so many historic and unrepeatable sounds.

This decision would seem to be justified by the fact that on more than one occasion broadcasts have had to be cancelled because the tape had been inadvertently wiped. Imagine if this had happened to an historic only copy! Most users of magnetic storage for sound, video and computer programs will know that this is all too easily done!

Video recordings in particular would seem to be vulnerable. Each picture field with all its colour information and light and shade detail is stored in a single micro-thin magnetic diagonal line, much narrower than a human hair, across the width of the tape which itself is a thin plastic film. Compared to that a cine film seems positively robust.

As most readers are aware, erasure of a magnetic recording is usually done with a magnetic field. In modern equipment, this is nearly always an alternating field applied either from an erasing head or from a special erasing unit. In some older cheap reel-to-reel recorders erasure was achieved by bringing a permanent magnet into contact with the tape.

Quite fascinating stories have circulated which tell of a credit-card company that was almost put out of business when a workman walked through its computer centre with a magnet in his toolbox, thus erasing all the magnetically stored data. An even more interesting one concerns the Inland Revenue records that were wiped clean by a nearby airport radar! Hope springs eternal . . .

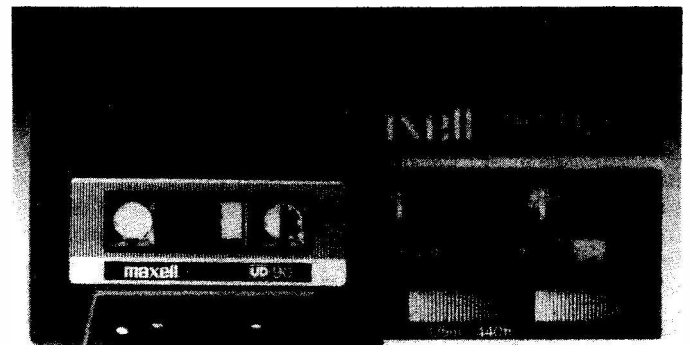
Another way that the magnetization of a tape can be destroyed is by heat. Apart from the effect of heat on the plastic substrate, above a certain temperature (known as the Curie point), a magnetic material will lose its ability to hold magnetization. This phenomenon is made use of in some brands of thermostatic soldering iron. Here, a disc made of magnetic material is placed in the bit of the iron. In the barrel of the iron is a small magnet, which, because it is attracted to the material in the bit, holds a contact shut. When the iron comes to the required temperature, which is also the Curie point of the bit, this attraction ceases, opening the contact and cutting off the heating current.

Now before everyone dashes out to dispose of their video recorders and floppy disc drives so as to get the best price before the rush starts, let us take a closer look at the

process of demagnetisation. A fully magnetised tape, like any other magnetised material, needs a certain minimum field strength applied to its surface to impress or remove magnetism. This field strength is known as the **coercivity** of the material. To understand how this works, you could think of a thug trying to get you to do something you didn't want to do — for example, hand over your money. If the thug applies enough coercing force, you must give way. However, your capacity to resist will depend on how strong you are. It's just the same with magnetic materials — some are much harder to magnetise (and to demagnetise) than others.

For audio cassettes a figure for coercivity of around 300 to 400 (ie 2.4 to 3.2 x 10⁴ A/m to our SI readers) oersteds is common, video tapes are usually somewhat higher. This means that a magnetic field of that order would be required to demagnetise a fully magnetised (or **saturated**) tape. Of course, tapes are not recorded into saturation otherwise the recorded signal would suffer distortion, so a normally recorded tape would be completely erased by a lesser field than that of the specified coercivity. Even so, it would take a field of at least 100 oersteds to do any damage to the recording.

So what sort of fields do we find around domestic equipment? External fields depend on the current flowing in the apparatus and the number of turns if a transformer, motor winding or other inductive component is involved. It also depends on the efficiency of the internal magnetic path. For example, a toroidal mains transformer is more efficient in containing the field through its core than a laminated type. Thus there is very little field **external** to a toroidal transformer. Equipment screening is another factor.



Magnetic tape — too ephemeral for the BBC.

A power drill running under full load will take a heavy current and the internal magnetic path is not particularly efficient taking in, as it must do, the rotating motor armature. So we can expect a sizeable external field. Surprisingly, the field at the casing of a domestic power drill under load has been measured at around 10 oersteds. This is well below that which could affect a recorded tape. House wiring and flexible mains leads also generate fields, but even when carrying a heavy current these are not great. The reason is that cable is reasonably straight and so constitutes only a single turn compared with the hundreds of turns of a transformer or motor windings. Further more, both live and neutral are contained within the same cable, and as these are carrying equal and opposite currents, there is a high degree of cancellation of the produced field. (For maximum cancellation the conductors would need to be tightly twisted.) So there is not much to worry about from these.

Permanent magnets are fairly common in domestic equipment — a few examples are in loudspeakers, tin openers, magnetic door catches, magnetic switches for burglar alarms, etc. Some of these produce fields of 1,000 oersteds and more, and could constitute a real hazard. Any such device coming into direct contact with a recorded tape or disc would certainly wipe that portion of the recording clean.

One factor which saves endangered recordings is that magnetic field is very strongly dependent on distance. For a single magnetic pole, the magnetic field would be proportional to the inverse square of the separating distance. However, magnetic poles come in opposite pairs, and this has the effect of making the field fall off even more quickly — the further you get away from the one pole, the more the field from the other pole tends to cancel out the field altogether.

This has the consequence that the casing around a video cassette, for example, will be sufficiently thick to protect it from contact with most small magnets — but it's still probably not a good idea to have a magnetic catch on the door of the cupboard in which you keep your tapes, just in case!

The situation is somewhat different with a floppy disc. This is contained in a protective packing similar to an ordinary record sleeve except that the disc is played inside the cover through a slot. The sleeve is there mainly to protect against finger marks and other minor handling hazards. It is not very thick, about 0.5 mm, so any magnet that was brought into contact with it would almost certainly wipe part of the information on the disc. Therefore, some care is needed in handling and storing these.

Then what about those stories of chain-reactions, with whole shelves of tapes being wiped out by a single stray field? It is possible for such a reaction to take place along the length of a single tape if the recording is of very low level. This was demonstrated some years ago by a team who were developing a method of recording very high frequency square waves with zero bias. The recorded signals were of very low amplitude but at one stage the recording level was accidentally increased to ten times that of normal. It was found that the previous 1,000 feet or so of recording had been wiped out by the excessive flux from the recording head at that one point travelling backward. To verify that this in fact was the cause, the conditions were repeated and other possible factors eliminated. The result was the same: erasure of the entire reel.

This however, does not happen under normal circumstances. In fact, it is a common practice to erase part of a tape by re-recording it, while the remainder stays unaffected. Audio tape enthusiasts having open reel machines often eliminate clicks and bumps in amateur

productions by judicious use of a permanent magnet, and post-recorded fades can be made by sweeping a magnet over the particular length of tape while gradually raising or lowering it.

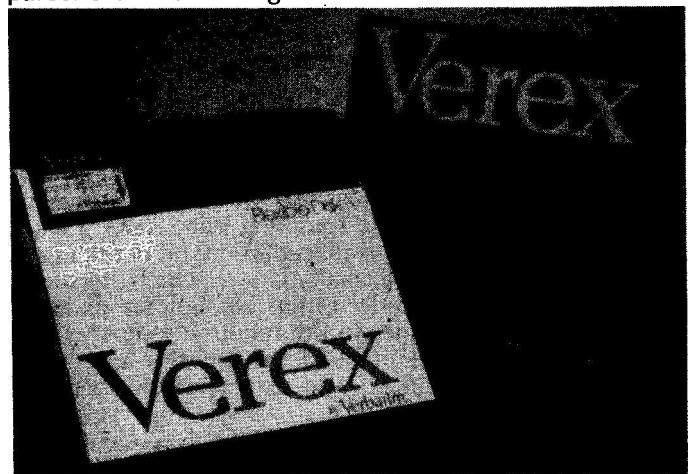
So with normal recordings the tape affected is only that portion which comes in close proximity to the magnetic field. As for several tapes being wiped by chain reaction, this is just a myth. The only way a number of tapes could be affected would be if there were a pervading field of such intensity as to exceed the tape coercivity at the surface of the tapes themselves.

There are stories in circulation about flash bulbs being triggered off by radar installations, and demolition teams that won't use electrical detonators nearby; and indeed, directly in front of a radar dish a magnetic field of several thousand oersted can exist. This would certainly be enough to wipe any magnetic tape clean, but any person in the vicinity would already be experiencing a few problems on their own account! In any case, airports and other users of radar don't like bodies getting in the way, so the antennae are usually mounted well away from people.

However, the question of whether or not a strong electromagnetic radiation field could cause erasure is worth exploring. 3M, the tape manufacturers conducted some experiments to determine this in America. The object was to discover the effect of microwave energy. Reels of recorded tape were actually placed in a microwave oven and the power applied until the reels and the tape began to melt and burn. Those parts of the tape that were not physically damaged were examined and the recording was found to be unaffected. The tape had not demagnetised.

A further experiment was tried using radar. Reels of recorded tape were placed directly in front of a radar dish having a range of 250 miles. Two lots of tape were used, one placed at a distance of 18 feet, and the other much closer at 16 inches. These were scanned by the radar beam for 16 minutes, then removed and examined. It might be anticipated that the nearer tapes would have suffered from the magnetic field, but in fact both lots were unaffected. There was no physical damage, and the recorded signal level was the same as before.

So, radar can be eliminated as a hazard for magnetic recordings, but there is another potential danger for tapes that are transported by air travellers, and also by air freight. Firstly, baggage, or packages that are sent through the post which for some reason may give rise to some suspicion, may be examined by means of X-rays. In other experiments by 3M, recorded tapes were exposed to X-rays of much higher level than those normally used for parcel examination. Again there was no ill effect and the



Floppy discs — the most vulnerable to demagnetisation.

recordings were intact, with no loss of signal level.

Secondly, at most air terminals, passengers are required to pass through a weapons detector with their hand luggage. These devices are magnetically operated, so what effect could they have on a recording? The majority are passive, that is they do not produce a magnetic field of their own but measure changes in the earth's magnetic field caused by ferrous objects. These pose no threat at all to recordings.

Other detectors are active. These produce a magnetic field in a doorway or other region through which the passenger must pass. Any metal object distorts the field, and the change is detected by instruments. Usually the field employed is quite low, in the region of 20 oersteds, which is insufficient to affect a magnetic recording. There are some detectors, though, that go up to 100 oersteds. Although below the coercivity of fully saturated tapes, they do constitute a hazard. While in most cases no damage may result, if there is reason to suspect that a high intensity field is in use it may be safer to advise the staff that you have recorded tapes, and request a visual examination instead.

Could, under normal circumstances, the temperature ever go sufficiently high to affect recordings? Certainly it's common experience that the temperature inside a car with its windows closed can soar to the unbearable — particularly in countries blessed with warmer climates than Britain. However, for iron oxide the Curie point is around 850°F (450°C) and physical damage to the tape substrate or case (or to the recorded itself) would occur long before this temperature was reached.

The Curie point for chromium dioxide is 250°F

(120°C) which, although still above boiling point and hence unlikely to be attained in a domestic environment, can still cause problems. This is because as the temperature rises towards the Curie point (remember that physical phenomena are governed by the absolute temperature, and although 50°C looks a long way from 120°C, the equivalents 323°K and 393°K look a lot closer) magnetic materials become much more susceptible to small magnetic fields. This principle was at one time used in the copying of videotapes: heat was used to transfer the magnetic pattern from the master tape to an intermediate medium and from that to the copies.

So one possible effect of excessive heat is print-through, ie signals from one layer of tape getting superimposed on an adjacent one to give pre- or post echo. This can and does happen at normal temperatures when thin (long playing) tapes are spooled up without playing, for long periods. High temperatures can therefore increase this risk. Print-through is also more likely in the presence of an external field which would be insufficient to erase the tape.

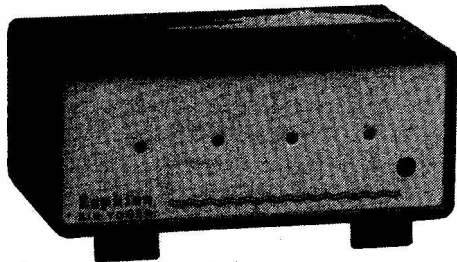
To sum up, there is not very much danger to magnetic tapes in most domestic environments apart from print through, provided the common sense precautions are taken. Obviously, if you allow a combination of hazard factors to occur together, the risk to the recording is increased commensurately.

I would add on a personal note that audio recordings made by me some 30 years ago have survived various hazards and sound as good today as they did on the day they were made.

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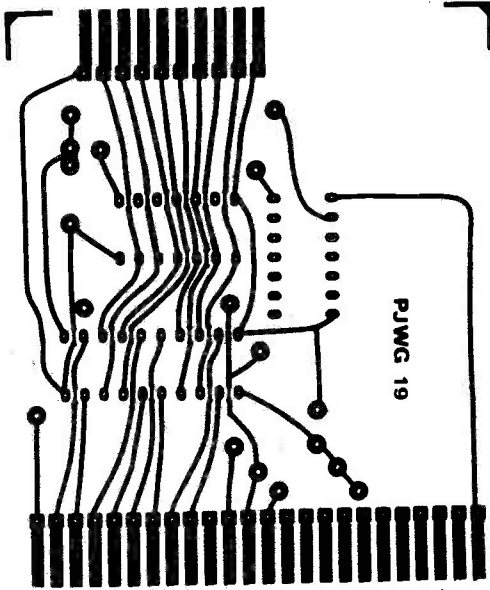
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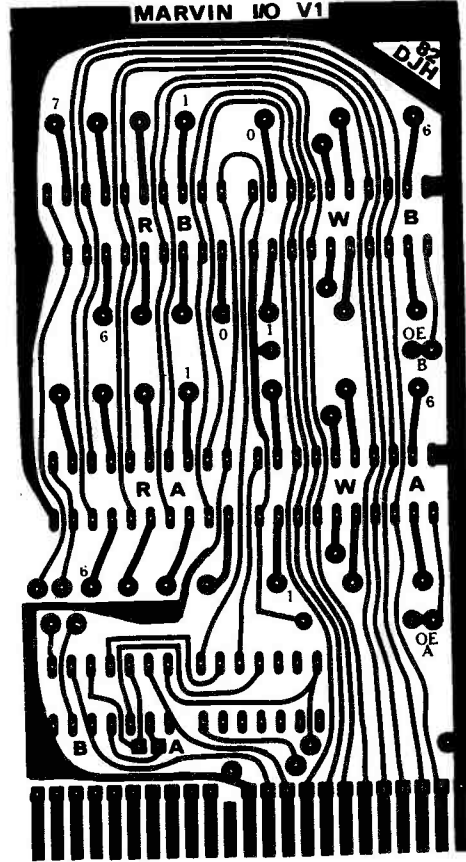
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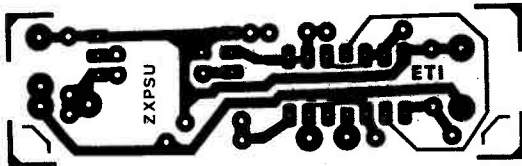
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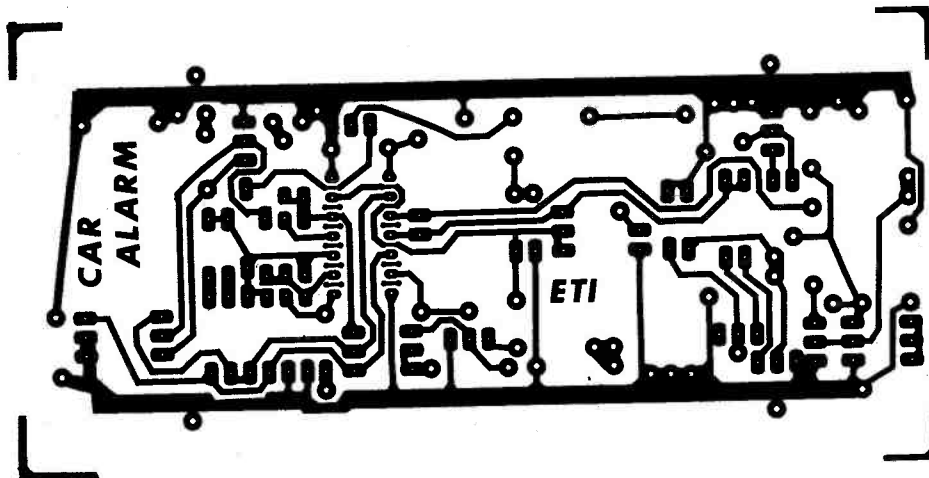
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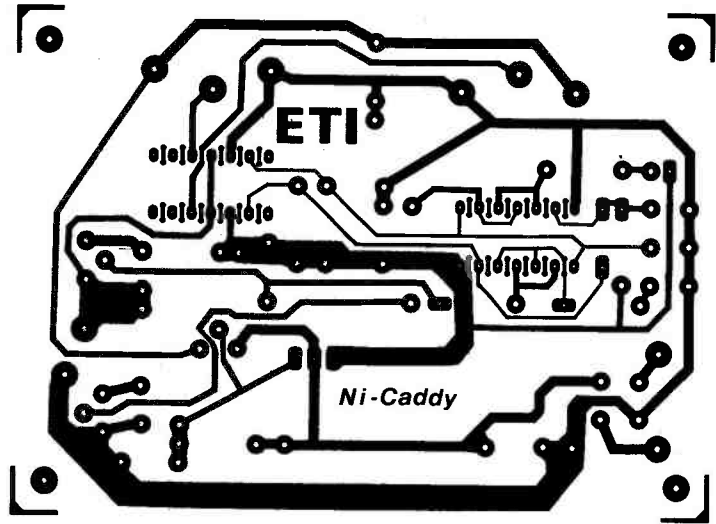
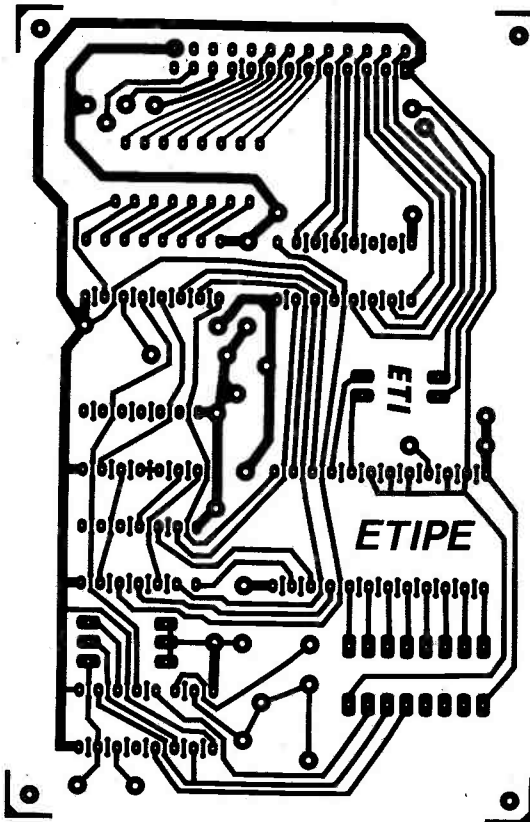


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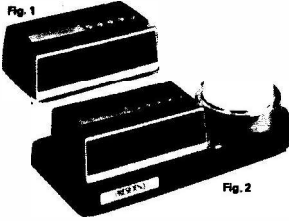
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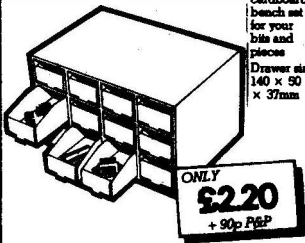
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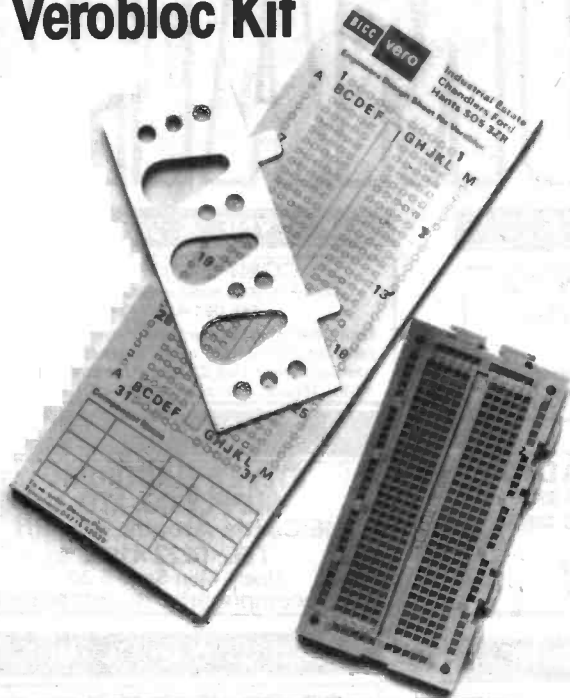
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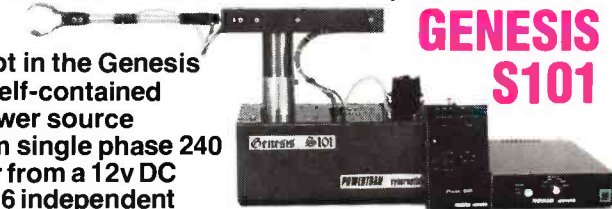
Aitken Bros	65
Ambit International	34
Audio Electronics	59
Bicc Vero	82
BK Electronics	5
Black Star	60
B.N.R. & E.S.	41, 60
Bradley Marshal	3
Clef Electronics	59
Comtech Electronics	16
Concept Electronics	65
Cricklewood Electronics	25, 26
Crimson Elektrik	52
Crofton Electronics	76
Dataplus Developments	52
Display Electronics	20
Electronize Design	27
Electrovalue	70
Frel Ltd	62
Greenbak	52
Greenweld	38
G.S.C.	49
Happy Memories	60
Hawk Electronics	60
Horizon Electronics	70
ICS	62
ILP	42, 43
Kelan Engineering	38
LB Electronics	62
L&B Electronics	65
L.E.M. Services	70
Magenta Electrovalue	70
Maplin	84
Marco Tradig	70
Mawson Assocs	70
Midwich Computers	19
MJL Systems	76
Musicraft	60
pantechnic	33
Parnden Electronics	76
Powertran	2, 18, 83
Rapid Electronics	12
Riscomp	33
R.T.V.C.	10
Sparkrite	8
Stuarts of Reading	59
Technomatic	54, 55
Thandar Electronics	19
TK Electronics	14
Watford Electronics	6, 7
W.E.B. Logic Systems Test Ltd	76

Low-price robots from POWERTRAN

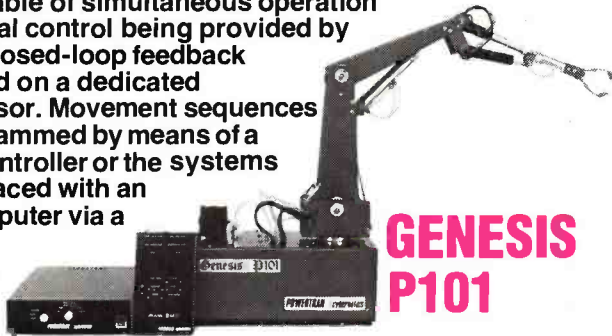
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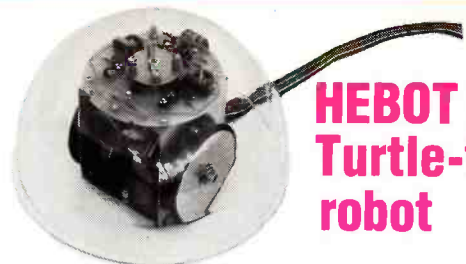
GENESIS P101

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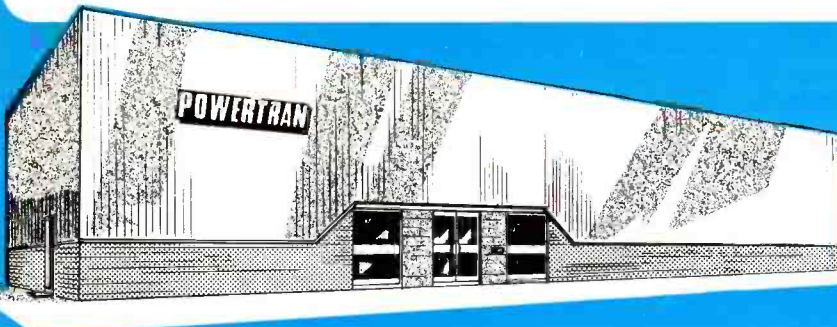
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*Projects for Book 8 were in an advanced state at the time of writing, but contents may change prior to publication (due 13th August 1983).

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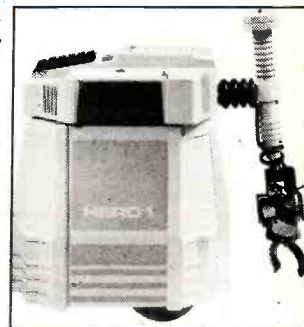
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