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| H6 | 40 | 250 mW ．Zener Diodes DO－7 Min．Glass Type | 50p |
| H17 | 20 | 3 amp．S．likon Stud Rectifiers． mixed volts | 50 p |
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| His | － 8 | Experimenters＇Fak of Incezrated Circuits．Date xupolied | 50p |
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| SEMICONDUCTORS |  |  |  |
| :---: | :---: | :---: | :---: |
| ACt07 | ${ }_{0}^{\text {app }}$ | OC139 | $<_{0.13}$ |
| ACl26 | 0.15 | OC140 | 0.15 |
| ACl27 | $0-17$ | －${ }^{\circ} 170$ | 0.21 |
| ACI2 | 0.15 | OC171 | 0.23 |
| AC176 | 0.20 | －C000 | 0.25 |
| ACYi7 | 0.20 | －c201 | 0.25 |
| AF239 | 0.30 | 2N1302－3 | 0.15 |
| AF186 | 0.20 | 2Ni304－5 | 0.17 |
| AF139 | 0.30 | 2N1306－7 | 0.20 |
| BCISA | 0． 20 | 2N1306－9 | 0.72 |
| ${ }^{\text {a }} 107$ | 0.10 | 2N3819FET | 0． 40 |
| eciou | 0.10 | 2N4416FET | 0.35 |
| ${ }^{\text {sclios }}$ | 0.10 | Powar |  |
| BCI 48 | $0 \cdot 10$ | Tramiutors |  |
| BC169 | 0.12 | － 20 | － 50 |
| $\mathrm{BFI}_{194}$ | 0.15 | Ocza | 0.10 |
| 87274 | 0.20 | OC25 | 0.25 |
| BFY50 | 0.15 | Ocrs | 0.25 |
| B5725 | －13 | Ocz | 0.30 |
| 85726 | 0.13 | 0 C 35 | 0.25 |
| BSY27 | 0.13 | $0^{0} 36$ | 0.37 |
| E5728 | 0.13 | AD149 | 0.10 |
| bsr29 | 0－13 | Aurio | 1.25 |
| ${ }^{\text {sfrysa }}$ | － 10 | 25034 | 0.25 |
| OC41 | 0.15 | 2N305S | 0.50 |
| 0 O44 | 0.13 | Dioder |  |
| $00^{45}$ | 0.10 | Aaym | 0.18 |
| OC71 | 0.10 | OA95 | 0.07 |
| 0 ch | $0 \cdot 10$ | OA79 | 0.07 |
| ${ }^{\circ} \mathrm{CB} 1$ | 0.13 | OAsi | 0.07 |
| $0 \mathrm{OCl} \mathrm{O}^{0}$ | 0.13 | OA95 | －． 07 |
| $0{ }^{0} 83$ | $0-18$ | in914 | 0.06 |

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$$
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& \text { 709C Linear Opp. Amp. } \\
& \text { Gates, Factory Marked and } \\
& \text { 25p } \\
& \text { Tested by A.E.I. } \\
& \text { 10p } \\
& \text { 10p }
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| :---: | :---: | :---: | :---: | :---: | :---: |
| 100\% | $0_{0.1}{ }^{\text {a }}$ | 6 p | 10v | 22 F | 7p |
| 100 v | $0-15 \mu 5$ | 6 D | 20v | $470 \mu \mathrm{~F}$ | 11 p |
| 1005 | 0-22 $\mu \mathrm{F}$ | $\mathrm{Bp}^{\text {p }}$ | 16\% | $47 \mu \mathrm{~F}$ | 75 |
| $100{ }^{\circ}$ | $0 \cdot 33 \mu \mathrm{~F}$ | 9 p | 25 v | $10 \mu \mathrm{~F}$ | 7 p |
| 100 F | $0 \cdot 4 \mathrm{~L} \mu \mathrm{P}$ | 10 p | 25v | $100 \mu \mathrm{~F}$ | 90 |
| $100 \%$ | $0.68 \mu \mathrm{~F}$ | 15p | 25 | $290 \mu \mathrm{~F}$ | 11 D |
| 2505 | $0.01 \mu \mathrm{~F}$ | 5 D | 235 | $470 \mu \mathrm{~F}$ | 14p |
| 2500 | 0-015 $\mu \mathrm{F}$ | 5 p | 255 | $1000 \mu \mathrm{~F}$ | 20 |
| ssor | $0-0223 \mu \mathrm{~F}$ | 5 D | 25 v | $2200 \mu \mathrm{~F}$ | 42 p |
| 2505 | $0-033 \mu \mathrm{~F}$ | 6 | 35. | $4.7 \mu \mathrm{~F}$ | 7p |
| 200 r | $0.037 \mu \mathrm{~F}$ | 6 D | 355 | $220 \mu \mathrm{~F}$ | 14 p |
| 250x | $0-068 \mu \mathrm{~F}$ | 6 p | 1000 | $10 \mu \mathrm{~F}$ | 8 p |
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ${ }^{\text {ACl27 }}$ | 197 | 160 | Series El? |  |  | 0 OL 70 | 249 | 210 |
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| ${ }_{\text {ACY18 }}$ | 18 p | 150 | SKT210 | 24 D | 19p | 184002 | 6p | 5 D |
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| AD162 | 27 | 95p | SKT? ${ }^{\text {S }}$ | 24 D | 19p | 184004 | 8p | 70 |
| AF139 | 289 | $28 \%$ | \KT213 | 260 | 19p | $1 \times 4005$ | 10p | 90 |
| BCl07 | 90 | Sp | \$KT?14 | 19p | 17 p | 1N4006 | 129 | 11p |
| BC108 | 8p | 78 | SET218 | 24 D | 19D | 174007 | 18p | 18p |
| BCl09 | 8 | 8p | \$KT219 | 24D | 19] | 184148 | 4 p | 3p |
| BCI 57 | 8p | 7 | 5LT233 | 29p | 20p | 217302 | 18p | 15 |
| BC148 | Sp | 78 | SKT\%en | $21 p$ | 19p | 2N1304 | 23 p | 80 |
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| :--- |
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## PROJECTS THEORY.

## SURREPTITIOUS SURVEILLANCE

The term "unlawful surreptitious surveillance" describes, most succinctly, the repugnant anti-social practice of snooping or eavesdropping upon private conversations with the aid of technical devices. This subject has received widespread attention recently following the publication of the report by the Committee On Privacy.

There is, however, another quite different form of "surreptitious surveillance" that one can perform quite legitimately and without giving offence to any innocent party. This is in order to protect one's property or premises from unionvied "visitors".

## BURGLAR ALARM

It is a regrettable fact that intruder surveillance systems are becoming increasingly necessary items of equipment for the ordinary home, just as for business premises.

There are of course many kinds of electronic burglar alarms in existence and they utilise various electronic properties in order to detect the presence of an intruder. Choice of a system is not always easy, and the environment in which it is required to function can play a large part in determining the effectiveness of a given system.

This month's burglar alarm design provides an inexpensive yet effective form of protection against the marauder. It exploits the sensitive-
ness of certain semiconductor devices to quite weak beams of invisible light (infra-red radiaion) and has a range adequate for many pourposes. Another practical application of simple electronics to meet a real and serious everyday need.

## EVENING CLASSES

Further proof of the spreading interest in d.i.y. electronics comes from the increasing number of non-vocational courses dealing with this subject, conducted up and down the country by local educational authorities. Many of these courses combine instruction in basic theory with practical demonstration, and also offer opportunities for students to build simple projects under expert guidance.

Evening courses start, generally, around midSeptember, so if you are interested in extending your activities in this way during the coming winter months make enquiries in your area without delay.

Local educational authorities who organise non-vocational courses for adults are usually responsive to genuine demands for specialist sub-jects-providing a suitable instructor is at hand.


Our October issue will be published on Friday, September 15

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## EASY TO CONSTRUCT SIMPLY EXPLAINED

VOL. I NO. II

SEPTEMBER
1972

## CONSTRUCTIONAL PROJECTS

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We are always interested in receiving readers' news and views of all items concerning this magazine and electronics in general; however if you write to us for advice would you please note the following.

We are unable to provide assistance on subjects not relating to published articles, and we cannot undertake to answer letters that do not include a s.a.e.


An alarm using an invisible, reflected beam to detect intruders.
By V. S. Evans
THis burglar alarm works on the principle that when a "dark light" beam is interrupted by a passing body, a power output is switched on for up to one minute duration. This output can be used as a direct power source for a small light bulb, or a relay to operate an alarm system.

OPERATION
A miniature bulb supplies the beam, which passes through a screen to filter out all the light spectrum except infra-red. Although the beam is now invisible it behaves like ordinary light and is focused by a lens, over the distance required, onto a mirror, where it is reflected back onto a second lens focused on a light sensor.



Fig. 1. Complete circuit diagram of the Infra-red Burglar Alarm.

The sensor generates current when the infrared beam is present, and this current is used to keep open an electronic switch. When the path of the beam is momentarily interrupted and cut off from the sensor by an interposing body, the sensor ceases to generate current and the electronic switch instantly closes, switching on power at the output, to operate the chosen alarm.

The switch remains closed, with the alarm operating, for a pre-determined period of time, then automatically opens and the system reverts to the "on guard" state.

## THE SENSOR

The beam sensor is a photo transistor (TR1) which is actually used as a photo diode, the collector not being connected. This device produces a current of 100 microamps or so when subjected to a simple light beam provided by LP1 as described above. This is sufficient to put a negative bias on the base of TR2 thus holding it in the "off" state until the current inducing beam is withdrawn (see Fig. 1).

Although the interruption may be for only a fraction of a second in duration it is sufficient for TR2 to momentarily switch on and deliver a pulse through Cl and D1 to the next stagethe monostable. Preset VR1 and R1 are used to set the level of standing current through TR2, so that the small change actioned by TR1 will switch TR2.

## THE MONOSTABLE

The monostable is one of a class of circuits known as multivibrators which change electronically between two states. The monostable, as its name implies, is normally dormant in one state.

If it is electronically activated into its second state it will remain changed for a pre-deter mined period of time and then automatically return to its dormant state.

The circuit of the basic monostable is shown in Fig. 2. It will be seen that TR3 is "on" due to a positive bias being applied to the base via R6. In this state its collector is near zero voltage, and through R4, applies this potential to the base of TR4, thus holding this transistor "off". In this state TR4 collector will be at near the positive line potential. It follows that capacitor C2 will have its positive side at near the positive line voltage and its negative side very much less so. It is therefore charged.

Now, if a positive pulse is delivered to the base of TR4, this transistor will switch "on," its collector voltage will drop to near zero and C2 will be forced to discharge at a rate controlled by the value of R6. During the period of discharge through R6, TR3 will be "off" and TR4 will remain "on". After discharge the circuit returns to its original state. This simple circuit can now be related to the circuit shown in Fig. 1. It will be seen that a variable resistor (VR2)

Fig. 2. Circuit to show the basic operation of a monostable.

has been included in series with R6 so that the time period of the monostable can be adjusted.

## POWER OUTPUT

In the circuit of Fig. 1, it will be seen that the emitter of TR4 feeds the base of the power output transistor TR5, consequently they work together, i.e. both "on" or both "off". The output transistor is a commonly available npn germanium type and should be up to grade with a low leakage rating, not a manufacturers reject or secondhand type. Resistor RI0 is purposely of low value to minimize any leakage that may nevertheless occur across the output. The diode D3 across the output is only required if the load is an inductive one, such as a relay or solenoid. It protects TR5 from reverse current caused at "switch off" by the field surrounding the coil, collapsing.

## FINAL CIRCUIT

Some of the components shown in Fig. 1 have not yet been explained; VRI is a preset resistor used as instructed later to set up the correct current in TR2. Diode D2 blocks the surge discharge from C2 applying a heavy reverse bias to TR3 base. Resistor R5 is the lower leg of the base bias potential divider to TR3. Diode D1 blocks any positive d.c. from interfering with the polarisation of C1. Resistor R7 and capacitor C3 form a filter which prevents any spurious pulse, which may be picked up inductively on the input line, from triggering the alarm.

## POWER SUPPLY

The unit will operate on a supply voltage from 9 to 12 volts, and this voltage will appear across the output when the alarm is triggered. For experimental purposes a large type of dry battery will suffice-as used for electric bells or lamps. For permanent installation an accumulator as used for scooters and cars is preferable. The unit will then operate even if the mains supply is off.

## SENSOR CONSTRUCTION

The beam and detector unit is shown in Fig. 3; this should be made up first. When completed this stage can be tested and put into operation. It should be solidly constructed and all parts firm. The dimensions shown are for the specified lenses. Infra-red gelatine sheet can be obtained from main photographic dealers or alternative materials can be found. Dark coloured polystyrene ${ }_{16}$ inch sheet has proved successful and the author has been told that resin bonded paper ( 0.015 inch Paxolin) can be used. However, the screen does not affect the working of the unit and can be left out until all testing and setting up has been done.

The sensor unit board is constructed as shown in Fig.4. Take care when soldering TR1, TR2 and

D1 and use a heat shunt on the wires being soldered. Make sure that VR1 is mounted in such a position that it can easily be adjusted when the board is fixed in the case.


## Photograph of the sensor circuit board.

Commence construction by drilling the board as shown and inserting the component wires through the holes.

The two Paxolin or s.r.b.p. insert panels, shown in Fig. 3, should be a tight fit, and are placed so that the bulb and photodiode are at the focal length from their respective lenses. This is the distance at which an image through the lens is at best definition on the panel and will be about $2^{3}{ }_{4}$ inches with the lenses specified. (The light from a window or a room light will provide the required image.)

The phototransistor has a light sensitive zone which must be located and placed at the front when mounting on the circuit panel. An initial location of this spot can be made by connecting the diode across a multimeter-say 500 microamp scale, and shining a pocket torch fairly close to and around the diode. This procedure will also show whether the device is a good one. The polarity is important when wiring up.

## ALIGNMENT

The unit has now to be aligned and this is best carried out in nearly dark conditions. Place it on a firm table or work bench and if possible it is best secured or clamped. The 12 volt supply is connected to the bulb only and the beam then directed at a flat surface 7 to 10 feet distant where the image of the bulb filament will show. Adjust the distance from lens to bulb for sharpest definition.

The beam should now be directed at a mirror (this can be as small as 1 inch square-Fig. 5 shows a suitable design) which can be angled so that the refiected light is made to cover the lens


Fig. 3: Construction of the beam and detector unit.



Photograph of the beam and detector unit.


Fig. 5. Basic design for an adjustable mirror.
focused on TR1. The phototransistor is then eased into a position which produces the highest reading on a meter connected across it, typically 100 to 150 microamps. With this achieved a few small drops of quick drying adhesive (clear Bostik or Uhu etc.) should be carefully placed each side of the diode to fix it.

## SENSOR TESTING

The next stage is to prove the functioning of TR2. With power connected (but not to the bulb) and a milliameter in the positive lead, adjust VR1 to give a reading of 1 to 1.5 milliamps. Connect the bulb directly to the power source (not through the meter) and if all is correctly lined up the reading should drop considerably.

Preset VR1 is then adjusted to give a standing current, in this state, of between 100 and 200 microamps. If the beam is now interrupted the reading should smartly rise to 1 mA .

## REMAINING CONSTRUCTION

With this part of the project working correctly, the construction of the second circuit board can be undertaken. This comprises the monostable and power output-wiring being straightforward as shown in Fig. 6.

The board is thin s.r.b.p. or Paxolin. All the components except the transistors and diodes should be mounted and wired up as shown. Once this has been done, carefully solder in the semiconductors using a heat shunt on each lead as


Fig. 6. Layout and wiring of the monostable and power output circuit board.



## Photograph of the alarm trigger unit.

it is soldered. Once again make sure the preset potentiometer (VR2) can be easily adjusted. Finally mount the board in a suitable case or on a supporting block. For test purposes a 12 volt bulb should be connected across the output. This should be a $2 \cdot 2$ watt 12 V type.

With power on, a 5 kilohm resistor placed briefly between the input and the positive lead, should bring the monostable into action and the bulb will light for a period, the length of which is controlled by the position of VR2. If a longer time period is required the value of $C 2$ can be increased, but there is a limit beyond which leakage can trigger the monostable and give false alarms.

It is important that at no time is the output short circuited as this would result in the output transistor overheating and probably destroying itself.

All being well the two units can be linked and proved functional. The main circuit board can be enclosed in any small box with the bulb or alarm mounted outside, or it can be enclosed with whatever alarm system the constructor chooses. A 9 or 12 volt relay (depending on the supply used) with a coil resistance of about 100 ohms or more can be wired up, as shown in Fig.7, to switch an alarm.

## USE

If the alarm is to be put into permanent use, it is essential that the beam/sensor unit and the reflecting mirror are fixed to solid supports and are absolutely rigid. Any slackness or movement will move the focal point of the beam off the

Fig. 7. Wiring used for a relay and alarm:

sensitive spot of the phototransistor, the resultbeing erratic performance and false alarms.

In installations of this kind it is sometimes recommended that screened cable should be used. In the prototype this was not found to be necessary and the connecting cables were cheap plastic covered wire twisted together for their full length of some 50 feet.

To provide protection around a room or building more than one mirror can be used to reflect the beam around the area as shown in Fig.8.

Fig. 8. Method of protecting an area.


## Components....

Resistors

| R1 | $100 \mathrm{k} \Omega$ | R6 | $47 \mathrm{k} \Omega$ |
| :--- | :---: | :--- | :---: |
| R2 | $5 \cdot 6 \mathrm{k} \Omega$ | R7 | $6 \cdot 8 \mathrm{k} \Omega$ |
| R3 | $3 \cdot 3 \mathrm{k} \Omega$ | R8 | $2 \cdot 2 \mathrm{k} \Omega$ |
| R4 | $15 \mathrm{k} \Omega$ | R9 | $680 \Omega$ |
| R5 | $56 \mathrm{k} \Omega$ | R10 | $47 \Omega$ |

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D3 1N4148 or similar
TR1 OCP71 germanium pnp
TR2 2N2926G silicon npn
TR3 2N2926G silicon npn
TR4 2N3704 silicon npn
TR5 AD161 germanium npn
Miscellaneous
LP1 Miniature 12V, 0.2A bulb and holder
VR1 $500 \mathrm{k} \Omega$ skeleton preset potentiometer
VR2 $100 \mathrm{k} \Omega$ skeleton preset potentiometer
LP2 $12 \mathrm{~V} 2 \cdot 2 \mathrm{~W}$ bulb (for alarm signal or testing only-see text)
RLA1 $100 \Omega$ or greater 12 or 9 V relay with one set of normally open contacts (to operate alarm system-bell etc. -see text)
Lenses-two jewellers eye glass magnifiers, one four way and one three way connection block, s.r.b.p. or Paxolin sheet (three pieces, 2 in $\times 2 \mathrm{in}$ ), materials for cases (see text), 4BA fixings, wire, small mirror and materials for mirror mounting (see text). Infra-red screen (see text)


Although satellites are providing a growing proportion of transatlantic communications links, the more down to earth system of undersea cables is undergoing exciting developments of its own.

Improvements in the technology of undersea cables-reduction in the loss per unit length of both the armoured and unarmoured cables and the use of semiconductors-are making a significant impact on the cost of laying transocean cables. For example, the Cantat 1 transatlantic cable was laid between Canada and Britain in 1961 at a cost of $£ 100,000$ a circuit, but a new cable, Cantat 2, to be laid in 1973-74 will cost just $£ 16,500$ a circuit.

The principal reason for such a dramatic improvement in economy is the increased number of circuits which the latest types of cable can carry. Cantat 1 has 80 circnits but Cantat 2 will have more than 1,800 circuits-more circuits, in fact, than all existing transatlantic cables combined.

## PAST AND PRESENT

Cantat 2 will be the third undersea cable between the UK and Canada. The first transatlantic telephone cable-TAT1 from Oban, Scotland to Clarenville, Newfoundland-was opened in 1956. Cantat 1 was laid between Oban
and Hampden, Newfoundland and was the first section of the Commonwealth cable network designed to carry calls between the UK and Canada and on to New Zealand, Australia and the Far East over Pacific and South East Asia cable systems.

Cantat 2 is primarily intended for Britain's communications with North America to meet a rising demand. Since Cantat 1 was laid the annual total of telephone calls from North America account for nearly $13{ }^{1}$ million minutes a year (compared to less than two million in 1960) and calls to North America from Britain occupy more than $10^{1}{ }^{2}$ million minutes (compared to $1^{1_{2}}$ million in 1960).

In addition to telephone calls, Cantat 2 will handle telex, telegrams and data transmission.

The existing transatlantic cables and the route of Cantat 2.



## TRANSISTORS

The key to Cantat 2 and the continuing future of long distance undersea cables as a practical and economic means of intercontinental communication lies in transistors which have been developed at the Post Office research establishment at Dollis Hill in North London. The transistors which replace thermionic valves, are able to operate with guaranteed reliability and performance to the higher bandwidths of submarine cable $-13 \cdot 7 \mathrm{MHz}$ in the case of Cantat 2 .

Built into the 2,840 nautical miles of Cantat 2 will be 473 repeaters each of which amplifies the signals and boosts them along their journey. Looking rather like torpedoes in their cylindrical steel housings and each weighing about a ton the repeaters will contain transistorised circuits which must have outstanding reliability.

In some places the cable will lie three miles deep and where pressure on the cable will be four tons per square inch. Apart from the cost of locating and raising the cable for repairs there would be a loss of operating revenue which, for Cantat 2 , will be some $£ 60,000$ an hour at full capacity.

Each of the 2,838 transistors contained in the cable has been designed to give a trouble-free life of more than 25 years. It is a standard of reliability that is unique and could be com-

This device creates the pressure condition the cable will encounter on the sea bed-up to three tons per square inch. The photograph shows part of the Cantat 2 cable being prepared for testing.

pared to switching on almost 500 transistor radios and expecting them to all work non-stop and perfectly for a quarter of a century.

## DESIGN ADVANCES

A further and simple indication of the advances made in undersea cable design in recent years is shown by comparing details of the Cantat 1 cable in 1961 with Cantat 2 with the latter's figures in brackets: number of speech channels $80(1,840)$; number of repeaters 90 (473); length in nautical miles $2,072(2,840)_{2}$ active elements 540 Post Office type 10P valves ( 2,838 Post Office type 4A and 10A transistors); principal types of cable, 0.99 inch unarmoured over 1,518 nautical miles ( 1.47 inch unarmoured over 2,425 nautical miles), 0.62 inch armoured over 554 nautical miles ( 1.47 inch armoured over 370 nautical miles); power, $9 \cdot 5 \mathrm{kV}$ and $415 \mathrm{~mA}(12 \cdot 34 \mathrm{kV}$ and 500 mA$)$.

Bringing Cantat 2 into service will cost about $£ 30,500,000$. This covers production, survey and development work and laying operations. The cost will be shared by the British Post Office and the Canadian Overseas Telecommunications Corporation but some rights of use will be sold to the authorised carriers in Europe and the USA.

The need to increase the gain-bandwidth, while at the same time keeping the amplifier voltage low, resulted in a change from thermionic valves in the Cantat 1 cable in 1961 to transistors for Cantat 2 which will become operational in 1974. This change precluded the use of parallel amplifiers which has improved the reliability of the early valve systems. Nevertheless the greater potential reliability of the transistor, compared to the hot cathode thermionic valve, more than compensated for the change.

In physical terms the gain-bandwidth product of the transistor increases as the transistor dimensions decrease. The trend is, therefore towards smaller devices.

## TRANSISTOR TYPES

The 4A type transistor developed by the Post Office and produced by both the Post Office and Standard Telephones and Cables Limited, provides 640 circuits in submarine use. However, the smaller types, known as the 10A2 and 10A10 designs, for input and output use respectively, allow an increase in circuit capacity to 1,840 -as in Cantat 2. Here the 4A type is used in the low frequency amplifier and the 10A types in the high frequency amplifier. This arrangement is also being used in a series of high capacity cables being laid in the North Sea linking Britain with Europe. The 10A2 was developed by the Post Office and the 10 A 10 jointly by the Post Office and STC.

The 10A type, being smaller, requires a more


A technician lifts a batch of transistor "headers"-tiny gold-plated beds on which the transistors will eventually rest-from an alcohol bath. This is part of a process which ensures that components are as clinically clean as possible. After being washed in alcohol, the "headers" are baked in a vacuum.
advanced technology than was needed for the 4A type. In particular, improved methods of diffusion have been developed for the 10A to give a base-width of 0.5 micron compared to the 4 A base width of 1.2 microns.

## RELIABILITY

The impressive reliability of the transistors stems from the method of bonding, by thermocompression, aluminium wires to aluminium contacts in each device. The standard of testing and inspection is such that of every 10,000 transistors made only 1,000 find their way into a cable system. The remainder are tested to destruction or do not meet the stringent standards required.

During the production of Cantat 2 some 20,000 transistors will be exhaustively tested. Already during the development of the transistors for submarine cable use 40,000 have been tested in production and 6,000 are in use in other cables on the sea bed. No failures have been found during these tests.

Although the provision of such reliability obviously becomes very expensive for trans-
ocean cable systems, the cost is more than balanced by the increased circuit capacity made possible by the improved design of solid state devices and which has reduced the cost per circuit nautical mile by a factor of 30 in 15 years.

A further aspect of the development work are the elaborate precautions taken to protect the transistors in the main amplifying path from the effects of electrical surges which occur if the cable itself is cut or damaged. This protection is provided by the use of diodes, which themselves must be highly reliable, to absorb the surges in both the power feed paths.

## CLEAN ATMOSPHERE

Apart from the advanced and highly skilled technology involved in producing the transistors themselves, the most important requirement is a perfectly clean atmosphere. Even the tiniest speck of dust will contaminate the transistor on which it settles.

The Post Office and STC, therefore, produce transistors and assemble the repeaters in "superclean" laboratory conditions where staff dress like surgeons. They often need microscopes to see their work and breathe the purest air. Sophisticated air conditioning systems filter and remove from the atmosphere the tiniest traces of dirt and dust.

So critical is the standard of cleanliness required that a member of the staff could create unwanted particles of dust simply by scratching his head. In fact the air in the laboratories is filtered and purified to such a degree that by comparison the air in a hospital operating theatre seems dirty.

## NAVIGATION

The precision which goes into the manufacture of the cable's components is continued when the cable is actually laid. To select the best possible route for the cable and to establish its location with pin-point accuracy should repairs be necessary, the survey vessels and the laying vessels must be able to navigate with extreme accuracy.
The normal methods of navigation used in commercial vessels do not meet these requirements and special arrangements are necessary; these largely involve the use of satellites.

Ships engaged in preliminary surveys and the actual lay will, therefore, use the Decca Hi-Fix system when covering the approaches to the terminals on both sides of the Atlantic-at Widemouth Bay, Cornwall and near Halifax, Nova Scotia. For the main part of the route across the Atlantic the ship's positions will be fixed by means of satellite navigation backed up by the Omega and Loran " C " navigation systems.

All these radio-navigational systems have a high standard of accuracy and are independent of weather conditions. Hi-Fix is a very accurate
short-range system with the twin virtues, for the cable ship, of repeatability and predictable accuracy.

The first, which ensures the ability to return to a previously visited point, is important from the maintenance aspect in an area, such as the approach to a cable terminal point, where existing cables are likely to be close together as they converge on the land station.
The second, predictable accuracy, allows the ship to be taken to a pre-determined point so that the cable can be laid along the route previously surveyed and where all abstacles have been mapped. This is vitally important where the cable, as in the case of Cantat 2, is routed along valleys and through passes when crossing the undersea "mountain range" known as the MidAtlantic Ridge. Hi-fix can only be used close to the shore otherwise the satellite system must be used.

Positions obtained by satellite navigation are extremely accurate but, at present, can only be obtained at varying intervals of time. The other radio-navigational aids, Loran "C" and Omega are not in themselves fully able to meet the exacting navigational demands of the project but provide a valuable back-up system. By careful observation they can be used to give information in the intervals between satellite fixes.

## CABLE LAYING

The increase in diameter and the closer repeater spacing needed in modern cables to handle the rise in circuit capacity, has led to

Continued on page 609

Throughout its life the cable will be subjected to incredible natural pressures, in shallow water the pull of tides and currents will create huge stresses. Additionally during laying and recovery for maintenance it is subjected to very heavy bending stresses. These two large rotating wheels simulate such stresses in the laboratory.


## A simple radio tuner for use with almost any amplifier. By F. C. Judd

THis very simple radio tuner will operate with a few feet of wire for an aerial and tunes to the medium and long wave bands. Sensitivity is sufficient to bring in local stations such as BBC Radio 1 and Radio 4 on medium wave and Radio 2 ( 200 metres) on long wave.

The tuner is ideal for tape recording and has an output of around 100 mV , depending on thestrength of received signals, and so can be directly coupled to the radio input of any tape recorder. It is suitable of course for use with any amplifier having an input sensitivity of around 100 to 200 mV .

Reception strength does depend on location and it would be unreasonable, for example, to expect strong signals from the 200 metre BBC long wave Radio 2 in remote parts of the country. On the other hand both sensitivity and selectivity do to some extent depend on the length of the aerial which may be any thin insulated wire about 10 to 15 feet long and as high in the room as possible.

If, for example, location is close to local stations such as the BBC Brookmans Park medium wave Radio 1 and 4, then a much shorter aerial would be needed to achieve complete separation of the signals. (The Brookmans Park station aerials are shown on the front cover and on this page.)

## THE CIRCUIT

As shown in Fig. 1 the circuit consists of a tuned radio frequency amplifier (TR1). The inductor $L 2$ provides a high resistance to the r.f. signals which pass through $\mathbf{C 5}$ to the diode detector. The main tuning capacitor (C2) is a


## SMedium 68 Long Wave

## Radio Tuner



Fig. 1. Complete circuit diagram of the LW/MW Radio Tuner.
mica dielectric type and the tuning coil a Repanco type DRR2 which has a tapped winding for medium and long wave tuning plus a secondary winding suitable for coupling to the low impedance input of TR1.
The output from the diode detector is taken to TR2 which operates as an audio signal amplifier. Radio frequency signals are removed by the capacitor C6, leaving the audio signal to be amplified by TR2.

## CONSTRUCTION

The prototype, as shown in the photographs, was constructed on a piece of plain circuit board and housed within a metal box. The box should be large enough to accommodate a PP6 type 9 V battery as well as the radio tuner circuit board, etc.

Details for the component board layout and wiring are given in Fig. 2 and the aerial socket, output signal socket and wiring to the remaining components are shown in Fig. 3. Note that the connections of the OC44 are located by the red spot (next to the collector lead) whilst those to the NKT 274 are according to position as in the inset of Fig: 2.

Commence construction by cutting the component board to size and drilling the board for the mounting bracket and L1. Fit LI to the board making sure it is the right way round (this makes the wiring neater). Next fit the remaining components, except the transistors and diodes and wire up as shown in Fig. 2. Check the layout and then wire in TR1, TR2 and D1 observing the correct connections.

Mount the circuit board on the front panel as shown together with the remaining components and wire up the complete tuner as shown in Fig. 3. We advise readers to follow the layout shown as alteration could cause instability.

Photograph showing the construction of the Radio Tuner. The aerial and output socket can be mounted on the side or back panels.


## CMedium 68 Zong Wave

## Radio Tuner



Fig. 2 Layout and wiring of the circiit board for the Radio Tuner.


Fig. 3. Connection of the circuit board to the remaining components. In the prototype the aerial and output sockets were mounted on the back of the case.

## SETTING UP

With a 9 V supply the tuner takes a little under 2 mA so if a milliammeter is available this could be checked to ensure correct operation of the two transistors. The DRR2 tuning coil may be supplied with a tuning core and if so, this must be removed completely; it is not needed. Couple up about 10 feet of insulated wire for an aerial and connect the output of the tuner to an amplifier, or to a tape recorder with through monitoring, so that signals are audible. The tuning points of local stations on medium waves will depend on location but those for the London area will be similar to those shown on the cover photograph.

If the received signals are strong they may overlap on medium waves. If this happens, reduce the length of the aerial until separation is obtained. No earth is necessary as the tuner will be automatically earthed via the amplifier or tape recorder to which it is connected.

## Components ....

Resistors

| R1 | $180 \mathrm{k} \Omega$ | SEE |
| :---: | :---: | :---: |
| R2 | $1 \mathrm{k} \Omega$ | ( 1 |
| R3 | 4.7k $\Omega$ | a) 1 |
| R4 | $10 \mathrm{k} \Omega$ |  |
| R5 | $100 \mathrm{k} \Omega$ | H |
| R6 | $4 \cdot 7 \mathrm{k} \Omega$ |  |
|  | W $\pm 10 \%$ carbon |  |
| Capac | itors |  |
| C1 | 47pF |  |
| C2 | 500pF variable (mica | dielecţric type) |
| C3 | $0.01 \mu \mathrm{~F}$ |  |
| C4 | $0.1 \mu \mathrm{~F}$ |  |
| C5 | 560pF |  |
| C6 | 4,700pF |  |
| C7 | $10 \mu \mathrm{~F}$ elect. 12 V |  |
| C8 | $10 \mu \mathrm{~F}$ elect. 12 V |  |
| C9 | $50 \mu \mathrm{~F}$ elect. 12 V |  |

## Semiconductors

TR1 OC44 germanium pnp
TR2 NKT 275 germanium pnp
D1 OA 91

## Miscellaneous

L1 Repanco type DRR2 (medium and long wave coil)
L2 Denco RFC 5 radio frequency choke (inductor)
SK1 Single insulated socket
SK2 Insulated phono socket
B1 9V PP6 battery
S1 S.p.d.t. slide or toggle switch
S2 S.p.s.t. slide or toggle switch
Metal case (or any suitable case, minimum size 4 in $\times 4$ in $\times$ 3in), tuning knob, aluminium angle $\frac{1}{2}$ in $\times \frac{1}{2}$ in $\times 2$ in (for mounting the circuit board), plain perforated Veroboard 3in x $2 \mathbf{z i n}$ $\times 0.15$ in matrix, material for tuning dial, 6 BA fixings, battery connectors.


Have you ever wanted to use your cassette recorder in the car or home and wished that you could power it from the car battery or mains supply? Well now you can, we will show you how to build two separate power supplies to cope with these requirements.

## HIECTRONIC MOUSE TRAP...

Why kill the mice that plague your home? This "humane" mouse trap catches the mice unharmed so that you can release them outside your home.

## Reactomailc...

Test your reflexes against other peoples! When the timed light comes on press your button first and your opponent is blocked-your light shows the winner. The Reactomatic can be developed for TV-type quiz game answering.

Next month's feature articicl... EIECTRONICS IN MEASUREMENT

All in the October issue.
On sale Friday, September 15

Inductors

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| :--- | :--- | :--- |


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$\overline{\operatorname{mon}}$ 4


## Meters



Winding of an inductor, coil, choke of transformer

Winding with a solid ferrite or dust iron core with a single tapping

Winding with laminated core

Winding with ferrite or dust iron gapped core

Winding with Faminated gapped core

Inductor with variable inductance, for example, tuning coil or solenoid with moving core

Transformer with two windings, a laminated core and a screen between the windings, no tappings

Auto transformer with one winding

## Ammeter or milliammeter

## Voltmeter or millivoltmeter



3


# symbols : - - part 4 



## (1)



78911



## Motors and Generators

G $G$
(M) H
(M) $\square$ M 13

$$
\text { (G) } 6
$$

(MS) $M$


M

Motorised switch or valve mechanism

## Rotating a.c. generator

## Rotating a.c. motor

## Synchiroñous motor

Squirrel cagedinduction mōtor


WITHOUT doubt, the most popular application of electronics is to produce audible sounds from a loudspeaker. In most instances, we hear the necessity of using an amplifier to produce the end product. However, it is not always so obvious why we need amplification or, indeed, what form it should take.

We have already produced quite a reasonable level of sound from a loudspeaker without recourse to amplification-i.e., the multivibrator in Part 8.

The reason why we needed no amplifier was because we had comparatively large voltage swings (9V) and reasonably high currents available in the collector circuit of the multivibrator. To produce the audible sound we had to feed these currents to a loudspeaker.

## POWER LEVELS

To obtain sufficient current we had to produce a high enough voltage swing to make this current flow through the impedance of the loudspeaker coil. It would be more correct to say we needed power to produce a sound of acceptable level.
The multivibrator was capable of delivering a power of about 125 mW to the 35 ohm loudspeaker. The power required by a loudspeaker to produce audible sounds depends on the efficiency of the loudspeaker and also on what we require in the way of volume.
We can describe the subjective effects of different power levels by saying that the output of a typical transistor portable radio would be
about 1W. The lowest power that will produce an easily recognisable sound (without putting the ear right up to the loudspeaker cone) would be about 10 mW and, of course, at the other extreme "pop" groups frequently delight the ear with powers greatly in excess of 50 W per amplifier.
To get maximum power dissipated in a loudspeaker we must apply the same reasoning that we arrived at in Part 4, i.e., the impedance (we are now dealing with alternating currents) of the circuit producing the current, should, as far as possible, equal the impedance of the loudspeaker. Most loudspeakers have very low impedances ranging typically from 3 to 35 ohm.

## POWER AMPLIFICATION

An amplifier serves two purposes. The first is most easily understood and is simply to increase the voltage swings that the source (e.g., microphone, pick-up, or radio tuner) produces. The second, and not so obvious role of an amplifier is to take these voltage swings and increase the amount of current produced.

We could say we are amplifying the current, although in practice we are using voltage swings in high impedance circuits, to produce equal voltage swings in low impedance circuits. Going from high to low impedance circuits without changing voltage levels, is the same as saying "current amplification"-going from low to high impedance is usually "voltage amplification." Combine the two together and one gets "power amplification."

## CRYSTAL MICROPHONE

Let's take the case of a crystal microphone. This is a device that converts air pressure waves into electrical voltages. The pressure waves impinge on a light diaphragm that is connected to a crystal made of "piezoelectric" material. This material has the property of producing minute voltages when it is mechanically deformed.

The level of signal produced by a normal voice about 6 inches from a crystal microphone is seldom much more than 10 mV . Unfortunately the piezoelectric crystal is a very poor conductor of electricity; so poor in fact that it behaves rather like a low value capacitor, the electrodes making contact to it acting as the plates.

Different types have widely varying characteristics but typically the capacitance is in the order of a few hundred picofarads. At 1 kHz this gives ann impedance greater than 100 kilohm.

The frequency range of the human voice is from about 100 Hz to about $3,000 \mathrm{~Hz}$, so you can see that at low voice frequencies the microphone has extremely high impedance, and as the frequency increases its impedance falls, but nevertheless is still very high and will probably be around 50,000 ohms.

## POWER IN LOAD

We run up against this "non constant" problem all the time in audio work so it is usual to talk in terms of a fixed frequency of 1 kHz unless otherwise stated.

Let's assume, then, that we have a microphone giving an output of 10 mV at 1 kHz and it has an impedance of 100 kilohm. What is the maximum power we can expect to obtain in the load "Z" shown in Fig. Ia?


Fig. 1(a). The capacitor and a.c. voltage source shown in the dotted box represents the crystal microphone. We say it is the "equivalent circuit'.

We will assume that the microphone is an alternating voltage source in series with its own capacitance. It is easier to understand if we convert this circuit to something more familiar, see Fig. lib.

The battery represents the voltage generated by the microphane ( 10 mV ), R1 is the impedance of the microphone ( 100 kilohm) and R2 is the impedance of the "ideal" load (equal to value


Fig. 1(b). The simple d.c. equivalent of Fig. 1(a) where R1 represents the impedance of the microphone, B1 its typical output voltage and R2 the ideal load for maximum power coupling,
of R1) for maximum power coupling. The power in the load will be

$$
\begin{aligned}
P & =\frac{V^{2}}{R_{2}} \\
& =\frac{0.005 \times 0.005}{100,000} \\
& =0.00000000025 \text { watts }
\end{aligned}
$$

It looks a bit better if we say 0.00025 microwatts, but still it is minute and if fed directly to a loudspeaker, would produce no reaction whatsoever.

## BIASING FOR A.C.

We shall use a transistor to give us voltage and current amplification at the same time. It will use the minute current available from the crystal microphone to control its collector current which, of course, should be higher.

The current from the microphone will be fed to the transistor's base/emitter circuit. The trouble is that the output from the microphone is an alternating current going both positive and negative about zero and typically it will not exceed 10 mV peak. We already know that we have to make the base of an npn transistor at least 600 mV positive w.r.t. the emitter before any base current is passed, therefore the output from the microphone will not have any effect at all when it is on a positive half cycle-let alone when it is on a negative half cycle. We have to put the transistor into a "partial" state of conduction before we consider the effect of the microphone. This we call "biasing."

We usually try to bias a transistor so that the current flowing through the collector load resistor causes the voltage at the collector to be about half the supply voltage. Any small variations in base current ultimately caused by signals from the microphone will then cause the voltage at the collector to move either more positive, or down towards zero, depending on whether the microphone is producing positive or negative half cycles.
The first thing we have to do is set the biasing so that the output of the collector is approximately half the supply-we call this the
"quiescent" condition. To make matters simple for ourselves let's assume we want a quiescent current of 0.5 mA to flow in the collector of TR1 in Fig. 2a (there are various factors which control the choice of this current, but we will ignore these at this point).


Fig. 2(a). If $h_{\text {FE }}$ for TRI was exactly 200, $V_{\text {out }}$ would be approximately +5 .Ovolts. In practice this is most unlikely as $h_{\text {FE }}$ can vary considerably from device to device.

To set the collector voltage at 4.5 V with 0.5 mA flowing, the drop across R1 must equally be 4.5 V , thus we can calculate the value of R1 using Ohm's law ( $\mathrm{V}=\mathrm{IR}$ ).

$$
R 1=\frac{4 \cdot 5}{0.0005}=9,000 \text { ohms }
$$

Let's say 10,000 ohms ( 10 kilohms) as the nearest convenient value. This will modify the voltage at the collector to +5 V (relative to the emitter line) but this is near enough for our purposes.

To pass a current of 0.5 mA in the collector circuit, we must pass a base current of $0.5 / h_{\mathrm{PE}}$ mA .

Using a BCl 08 the $h_{\mathrm{FE}}$ will be around 200 , therefore $I_{\mathrm{b}}$ will be 0.0025 mA -we must provide this through our bias circuit.

If we use the positive rail as the source of base current we must limit it through a resistor (R2) and this will have a value given by 9 V minus 0.6 V , divided by the base current ( $I_{b}$ ). Therefore,

$$
\mathrm{R} 2=\frac{9-0.6}{0.0000025}=3,360,000 \text { ohms (say } 3 \cdot 3 \text { megohm) }
$$

## EXPERIMENTAL CHECK

Wire up this circuit on the Demo Deck and measure the potential at TR1 collector. The chances are that it will not be the +5 V as calculated because we have made the assumption that $h_{\mathrm{Fe}}$ was 200 , and this is not necessarily the case as it varies considerably from one device to another (use a $20,000 \mathrm{ohm}$ per volt voltmeter for this and other measurements).

We can "cheat" a bit at this stage and adjust the quiescent base current to give us the output level we require as shown in Fig. 2b. Wire up this circuit on the Demo Deck, VR1 is the 100 ohm potentiometer of the Demo Deck which we will use to set the drive voltage for the base


Fig. 2(b). Variations in $h_{\text {FE }}$ can be overc̄ōme by adjusting the bias with VR1 until the quiescent output voltage is +4.5 V .
current, and R2 is now made 1 megohm. Adjust the setting of VR1 until the collector voltage is exactly midway ( +4.5 V ).

Having set the d.c. state of our circuit we can now inject the signal from our microphone. Connect the pair of wires from the microphone between base and emitter. This should not affect the 4.5 V at the collector because the microphone is virtually a capacitor and will not modify the d.c. current in the base circuit. Any signal from the microphone will now add to or subtract from the standing base current.

Try speaking close to the microphone and you might just see a slight flicker on your voltmeter; probably not much because you are trying to measure a high frequency voltage on top of a reasonably high d.c. level. We can get rid of the d.c. level and at the same time rectify the alternating current by the complete circuit shown in Fig. 3.


Fig. 3. Complete common emitter amplifier stage with output voltage metering circuit.

Capacitor C 1 will pass the a.c. while blocking d.c. and diode D1 shorts out negative half cycles. You can now set your meter to a lower d.c. range and when you talk fairly loudly and close to the microphone, you should see voltage swings of about 1 volt.

Obviously we have amplified the approximate 10 mV output of the microphone, but what current swings do we now get? The IV swings are occurring across a 10 kilohm resistor, therefore the current must be varying by about $0 \cdot 1 \mathrm{~mA}$. Thus the power in R1 is being varied by at least $1 \mathrm{~V} \times 0.1 \mathrm{~mA}=0.1 \mathrm{~mW}$ and this is caused purely by the current injected by the microphone.

We still cannot apply 0.1 milliwatts to a loudspeaker and expect to hear anything, but at least we are talking about a level only one hundredth of the minimum desirable level and this is a vast improvement compared with the fractions of microwatts we have been talking about previously.

## GROUNDED EMITTER

The circuit we have made is called a grounded emitter amplifier stage because the emitter is connected directly to the negative power rail. Sometimes it is called a common emitter stage because the input source used the emitter line as one of the connection points, and we measured the output relative to the same emitter line.
The method of biasing is somewhat unconventional, but in this case is used to demonstrate the principle involved. We will later come across some more sophisticated ways of biasing.

## HIGHER POWER

Now let's press on and see if we can produce sufficient power from the microphone to drive a loudspeaker. We now have a signal level of 1 volt at low current. Assume we had unlimited current available; 1 volt across a loudseaker of 35 ohms impedance (as is that in the Demo Deck) would dissipate a power of

$$
\frac{V^{2}}{R}=\frac{1 \times 1}{35}=29 \mathrm{~mW} \text { (approximately) }
$$

This would be ample to produce an audible sound. Therefore 1 volt is sufficient, but we need more current. Now we shall use another transistor in a current amplifier circuit that does not change the voltage swings.

The simplest circuit to do this is called an emitter follower or grounded collector stage. The basic circuit is shown in Fig. 4a.


Fig, 4(a). Basic emitter follower circuit.
Notice that the 100 ohm resistor ( $\mathrm{R} \overline{3}$ ) is in the emitter circuit. In the absence of any base current, the potential at the emitter will be zero. We can now do another simple experiment on the Demo Deck using the circuit diagram of Fig. 4b.


Fig. 4(b). Experimental circuit to show the working of an emitter follower stage. Measure voltages at $B$ for different settings of voltages at $A$.

Connect a 10 kilohm resistor to the base of TR2 and take the other end to a 100 ohm poten-tiometer-used as a potential divider. Monitor the voltage at the emitter of TR2 and slowly increase the voltage at the wiper of VR1. At each setting of VR1 you should find that the potential at the emitter is the same, less about 600 mV .

The reason for this-is that base current is drawn by the transistor as soon as the base becomes 600 mV more positive than the emitter, but this base current causes collector current to flow and this causes the potential at the emitter to rise. The ratio of base to collector current (which is almost the same as emitter current) is again $h_{\text {Pe }}$.

The voltage at the emitter will rise until it nears 600 mV below the base voltage, and then the rise will stop; the emitter cannot rise more positive than the base-or even reach the same value-because if it did, base current would cease to flow, and hence the collector current would fall. Thus apart from the initial 600 mV difference, we say the emitter voltage "follows" the base voltage.

There is an important difference though in these two voltages. The one at the base is produced through a 10,000 ohm resistor, while that at the emitter is across 100 ohms. Notice we have in effect reduced circuit resistance (or impedance) across the transistor.

A voltage causing a low current to flow through a high resistor in the base circuit causes a similar voltage to appear across a much lower resistor, hence the current must be much higher. We can work out what the maximum current we can control in the emitter circuit will be by simply multiplying the available base current by $h_{\text {FE }}$.

## TWO-STAGE AMPLIFIER

Instead of connecting the base of TR2 to the wiper of a potentiometer through a resistor, we will connect it straight to the collector of TR1, see Fig. 5. All the voltage measurements we saw before will appear at the emitter of TR2 apart


from a 600 mV constant drop, but the currents flowing will be considerably higher.

A small proportion of the current flowing through RI is sufficient to provide the base current for TR2 without affecting the collector levels of TR1 too much.

Wire up the circuit of Fig. 5a on the Demo Deck (Fig. 5b). The 35 ohm loudspeaker of the Demo Deck is connected through a $500 \mu \mathrm{~F}$ capacitor. The potential at the emitter will be about $+3 \cdot 9 \mathrm{~V}$ (caused by the quiescent potential at the collector of TR1).

In the absence of signal from the microphone no d.c. will flow through the capacitor and loudspeaker, but if we speak into the microphone the fluctuations in emitter current will pass through the capacitor into the coil of the loudspeaker and produce quite a reasonable audio output.

The capacitor presents very low impedance to the path of a.c. and the small a.c. currents (compared with the quiescent current) are fed directly to the loudspeaker with only a small proportion being shunted by R 3 .

You will probably find that the circuit is so. sensitive that you will encounter acoustic feed-

Fig. 5(a) (below). Complete microphone amplifier giving a reasonable output into a 35 ohm loudspeaker. The effect of excessive or insufficient bias current can be experienced by adjusting VR1.
Fig. 5(b) (right). The microphone amplifier of Fig. 5(a) wired up on the Demo Deck.
back. This shows itself in the form of a "howl." It is caused by the sound from the speaker being picked up by the microphone, being re-amplified and fed back to the speaker-only to repeat the same cycle over and over again. The best way to prevent this is to separate the microphone and the loudspeaker by a reasonable distance.

## VARYING THE BIAS

Remember the bias is still set by VRI right at the front end. Try varying the bias in both extremes. By reducing the bias current to zero, you will find that the gain of the system reduces to zero; by increasing it you will notice, first of all, distortion which gets worse and worse until there is again virtually no output.


Photograph of the microphone amplifier connected up on the Demo Deck.

Everyday Electronics, September 1972

The former is caused by the first transistor refusing to conduct at all, while the second is due to the signals from the microphone driving the transistor into saturation until the bias itself makes the transistor fully conducting all the time. Try experimenting with different values of R1, R2 and R3 and see if you can arrive at any deductions regarding output levels or biasing levels. Do not make these resistors less than the following values: R1 1 kilohm, R2 10 kilohm, R3 100 ohms.

## FREQUENCY RESPONSE

Remember we said that the irfopedance of the microphone increased for low frequencies? This means that the amount of base current it can supply into TR1 will decrease for low frequencies and increase for high frequencies. Try "crooning" into the microphone and then whistling a high note (both at about the same volume) and you will hear that the output from the loudspeaker is very much greater for the higher whistled note. You can, in fact, measure the differences in amplitude if you go back to the metered experiment shown in Fig. 3.

We say that this amplifier does not have a "flat response" and hence does not reproduce the original signal to perfection-the fidelity is therefore poor.

In the case of, say, an intercom, this does not matter but it is a terrible fault to have if we are trying to obtain hi-fi. In quality designs,
steps are taken to reduce this effeot of amplitude "roll-off" at low frequencies.

Next month we shall show you two better ways of biasing the first transistor that will do away with the necessity for VR1 and also a better impedance matching stage that will give us more power output at higher efficiency.

Next month: Amplification. Components required in addition to those already obtained: resistors, $2.2 \mathrm{k} \Omega, 22 \mathrm{k} \Omega$, $150 \mathrm{k} \Omega, 470 \mathrm{k} \Omega, 560 \mathrm{k} \Omega$, $1 M \Omega$ (all $\frac{1}{4} \mathrm{~W} \pm 10 \%$, 1 off each). Capacitors, $2,000 \mathrm{pF}$ ( 1 off ), $50 \mu \mathrm{~F}$ elect. 15 V ( 2 off). Transistors, BC 108, 2N3702 (I off each).



# Ruminations BySensor 

## Other People's Jobs

While waiting to turn into a main road the other day, my car came to rest opposite one of those Post Office Telephones "tents" erected over a hole in the pavement. I peered into the tent just as a man climbed out of the hole; our eyes met. "Not a bad day for camping," I said, he considered my remark for a while then agreed, good naturedly, that it wasn't at all bad.

Later, I wondered what I would have said if he had leaned over my shoulder in the electronics lab and said, "Not much on the telly today," as I gazed into the oscilloscope. I hope that I would have been as agreeable as he was.

Other peoples jobs fascinate me. Fortunately, I find that most people are willing to talk about their work if they believe that the enquirer is really interested. Sometimes, if pressures of work are not too great, one may be invited to "have a go" and the experience can be most satisfying. I enjoyed, particularly, an opportunity to try my hand (and mouth) at glass blowing.

## A Little Knowledge

One frosty morning, the sink fitted in the electronics laboratory in which I was working, became blocked. By the time we had all rinsed our tea cups the water was an inch or so deep, and eager to play the plumber, I unscrewed the plug from the trap. The water draigied into the bin that I had placed beneath, but the sink was found to be still blocked when the plug was replaced. I admitted defeat and returned to my 'scope and "breadboard".

Later, someone else decided to
"have a go". He connected a hose from the compressed air supply to the sink outlet and turned on the compressed air-the water disappeared and did not return. "Well done," we said, and he took up his soldering iron again with a happy smile. A few minutes later, a man from the lab above came in. "You ought to have been in our lab, just now," he said. "A great jet of water shot out of the sink and went all over the ceiling!"

We expressed our sympathy, and the right amount of incredulity, and sent for the plumber.

Electronics engineers are a pretty dedicated lot on the whole, but like everyone else they enjoy a bit of fun. On one occasion we noticed a wire dangling from the lab above, "Ah, someone is building a radio" we said. We fished the wire through our window and connected it via a $0 \cdot 1 \mu \mathrm{~F}$ capacitor to a signal generator, while someone from our lab went up to "help" them to cure their oscillating radio!



Fig. 2. The complete circuit diagram of the Capacitance Meter.


Fig. 1. The basic circuit of the Capacitance Meter.
Approximate cost of components $£ 5 \cdot 50$ plus case
will restrict the largest measurable value to 3 microfarad.

## MULTIVIBRATOR SWITCH

The electronic switch used in the Capacitance Meter is an astable multivibrator. Looking at the complete circuit, Fig. 2, the multivibrator switching frequency is determined by switch


## CAPACITRACE

 metir

Fig. 3. The complete wiring diagram of the Capacitance Meter. Everything is connected to the back of the front panel.

Fig. 4. Suggested dimensions for the front panel, made from Paxolin, Formica or Perspex.
selected pairs of capacitors (Sla, Slb), switched base resistors ( S 2 ), and by an adjustable voltage applied to the base resistors (VR1, VR2).

With S2 closed, Sl gives decadal frequency steps from approximately 10 Hz to 100 kHz , and with S2 open 3 Hz to 30 kHz , the latter being the times three ranges.

Preset potentiometer VRI provides a fine adjustment of frequency for both S 2 settings, and thus acts as an overall calibration control to cater for falling battery voltage, while VR2 serves only for initial calibration of the times three multiplier.

## A.C. METER

The a.c. reading meter, made from $\mathrm{R} 2, \mathrm{R} 3, \mathrm{C} 6$, C7, MEI, DI, and D2 (see Fig. 2) is connected in series with the unknown capacitor $C_{x}$ between the collectors of TR1 and TR2.

Because of the steep sided exponential waveform fed by $C_{x}$ to the rectifier diodes D1 and D2, and a multivibrator output of more than $15 V$ peak to peak, errors due to diode nonlinearity are small. Tests with a large $\pm 1$ per cent meter showed a non-linearity of less than $0-5$ per cent over 98 per cent of the scale.
Battery drain of the circuit is only 5 mA , and the push to read button S3 will ensure that the meter is not accidentally left on after use, so battery life will be almost as good as shelflife.

## CONSTRUCTION

Fig. 3 gives drilling details and dimensions of a 7 in $x$ 5in s.r.b.p., Formica or Perspex front panel. A metal panel is avoided because it would tend to increase stray capacitance on the $0-100$ picofarad range.



Fig. 5. The layout of the components on the top side of the Veroboard with flying lead connections. Below is shown the regions of copper strip to be removed from the underside.

After lettering the front panel, mount VR1, S2, S3, ME1, SK1, and SK2, as shown in the general wiring and layout diagram shown in Fig. 4.

Switch S1 should be pre-assembled with capacitors $\mathrm{Cl}-\mathrm{C} 5$ and $\mathrm{C} 8-\mathrm{Cl} 2$ before mounting on the panel.

Next, solder the components to a $2 \cdot 1$ in $x 1 \cdot 4$ in piece of 0. lin matrix Veroboard, as shown in Fig. 5. The Veroboard cut-outs at positions H17, G17, 18, 19, 20 and 21, have been made to minimise stray capacitance between the copper strips.


Photograph of the top side of the Verobeard with all components in position.

The transistors and diodes should be the last components to be soldered in position and a heat shunt must be used on the leads when soldering, otherwise permanent damage may be done to these components.

Attach lengths of 22 s.w.g. tinned copper wire to the ends of the panel to form leads.

When the circuit panel is complete, offer it up to the front panel as close as possible to SK1 and SK2, and then cut the leads, insulate with sleeving, and solder the 22 s.w.g. leads to the front panel components.

The wiring on the back of the front panel of the prototype.


When this wiring is completed it will be found that the circuit panel is held quite firmly and will need no additional support.

Complete the wiring by interconnecting the front panel components, not forgetting Cl 3 .

## CALIBRATION

To check that the instrument is working on all ranges, temporarily connect a 33 kilohm resistor in series with a capacitor of about 0.5 microfarad between SK1 and SK2, and press S3. A meter reading should be obtained at all range settings.


The prototype in use.

Two silver mica capacitors of $\pm 1$ per cent tolerance, a 100 picofarad and a 10,000 picofarad, will serve to calibrate the meter. First connect the 10,000 picofarad standard to SK1 and SK2, set S1 to 0.01 microfarad and S2 to times one, and press S3; adjust VR1 for a full scale reading.
To calibrate the times three range with the same standard capacitor, rotate VR2 on the circuit panel fully clockwise, set S2 to times three, and then adjust VR2 carefully until the meter reads 0.01 microfarad when $\$ 3$ is pressed.

Return S2 to the times one position and set Sl to 100 picofarad. Place the 100 picofarad standard capacitor across SK1 and SK2 and trim Cl for a full scale reading. Remove the 100 picofarad standard capacitor and observe the residual stray capacitance reading, this should be no more than 3 picofarad.

Now adjust Cl again while measuring the 100 picofarad standard, to make the meter read 100 picofarad plus the stray capacitance, that is, slightly more than full scale.

When using the 100 picofarad range, always deduct the stray capacitance value from the indicated value to obtain the true value.
If it is found that there is excessive pointer vibration with the particular meter movement
used for ME1, on the 3 microfarad range, a capacitor of about $300-500$ microfarads can be wired across the meter terminals in series with a switch, to give additional smoothing on this range.

The capacitance meter is only suitable for measuring non-polarised capacitors with a rating of 15 V or more. A shorted capacitor will show up as a more than full scale reading on all ranges, while an open circuit component will give no reading at all.

## Components....

## Resistors

| R1 | $2 \cdot 7 \mathrm{k} \Omega$ |
| :--- | :--- |
| R2 | $4 \cdot 7 \mathrm{ks} \Omega$ |
| R3 | $4 \cdot 7 \mathrm{k} \Omega$ |
| R4 | $100 \mathrm{k} \Omega$ |
| R5 | $68 \mathrm{k} \Omega$ |
| R6 | $100 \mathrm{k} \Omega$ |
| R7 | 68 ks 2 |
| R8 | $2 \cdot 7 \mathrm{k} \Omega$ |
| All $\frac{1}{2}$ watt $\pm 10 \%$ carbon |  |

Potentiometers
VR1 $5 \mathrm{k} \Omega$ linear carbon, T.V. preset type
VR2 $5 \mathrm{k} \Omega$ miniature preset, horizontal mounting

## Capacitors

C1 200 pF compression trimmer
C2 $1,000 \mathrm{pF} \div 1 \%$ silver mica
C3 $\quad 10,000 \mathrm{pF}+1 \%$ silver mica
C4 $\quad 0 \cdot 1 \mu \mathrm{~F}$ polycarbonate or polyester $\pm 10 \%$ or better
C5 $1 \mu \mathrm{~F}$ polycarbonate, tantalum, or polyester $110 \%$ or better
C6 $0.01 \mu \mathrm{~F}$ polyester $\pm 20 \%$
C7 $100 \mu \mathrm{~F}$ elect. 15 V
C8 $\quad 100 \mathrm{pF} \perp 1 \%$ silver mica
C9 $1,000 \mathrm{pF}=1 \%$ silver mica
C10 $10,000 \mathrm{pF} \pm 1 \%$ silver mica
C11 $\quad 0 \cdot 1 \mu \mathrm{~F}$ polycarbonate or polyester $\pm 10 \%$ or better
C12 $1 \mu \mathrm{~F}$ polycarbonate, tantalum or polyester $10 \%$ or better
$\mathrm{C} 130 \cdot 25 \mu \mathrm{~F}$ polyester $+20 \%$
Transistors
TR1, TR2 BC 109 silicon npn (2 off)
Diodes
D1, D2 OA81

## Switches

S1 Two-pole six-way wafer
S2 Double-pole changeover toggle
S3 Single-pole push button
Miscellaneous
ME1 100 1 A (with 0-3, $0-10$ scales) $\pm 2 \%$ or better
B1 9V battery PP3
4 mm plugs and sockets; 7 in $\times 5$ in s.r.b.p., Perspex or Farmica front panel; 2.1 in $\times 1$ - 4 in Veroboard 0.1 in matrix; knob; 22 s.w.g. tinned copper wire; two crocodile clips.

Continued from page 591
further advances being made in cable laying methods. This includes a new type of cable laying "engine" designed by the Post Office. The new engine was first used to lay the UK-Spain cable in 1970.

A modified design of the engine will be fitted to the Canadian cable ship and ice breaker John Cabot which will be used to bury the Cantat 2 cable and repeaters in the shallow waters of less than 300 fathoms off the Nova Scotia coast to protect the cable from damage by trawlers.

The cable and repeaters are laid through a plough-type device and the John Cabot is at present the only ship available powerful enough, because of her other activity as an ice breaker, to pull the plough. Remote TV cameras are fitted to the plough and linked to monitors in a control room on board. More TV monitors and indicators showing cable tensions and ground speed of the plough are also mounted on the ship's bridge.
The plough is capable of cutting a furrow in the sea floor allowing the cable to be buried to a depth of up to 26 inches. The other ship which will be engaged in laying 95 per cent of the cable is the Cable and Wireless ship, Mercury, which is also to be fitted with a linear laying engine.

After all the work of the electronic experts in developing today's trans-ocean cables and the use of the latest scientific navigation and monitoring devices nature must still be accounted for. Storms are still a hazard to the cable ships. Photographs taken from the bridge of a Post Office cable ship are used by the Meteorological Office as a standard reference to illustrate sea conditions in a hurricane.

## ADDENDUM

Since this article was written the Post Office's Research Department at Dollis Hill has received the Queen's Award for Industry, given for "Technological Innovation" in the development and production of high quality transistors for use in undersea telephone cables.


Cable Ship Alert lying to final splice at Kennack Sands, Cornwall during the laying of a U.K.Spain cable.


Anumber of readers have written to us regarding the supply and use of Veroboard, since we feel that there may be a great number of readers who are not fully aware of the types, we will try and clarify the situation.
Firstly, there are three fairly common types and we use the following terms to describe them; Veroboard-by this single name we mean perforated board with lines of copper on one side only, this is probably the most common and the one we usually use; plain Veroboard-this is perforated board with no copper strips on it; double sided Veroboard - perforated board with copper strips on both sides, we have never yet used this type of board and it is doubtful if we ever will, as it is not necessary for the type of projects we describe.
We usually use 0.15 inch matrix, this means that the rows of holes are 0.15 inches apart, there are other sizes ( 0.1 inch and 0.2 inch) so make sure you get the right one.

## Infra-Red Burglar Alarm

The main buying problems for the Burglar Alarm are more likely to concern the non electronic parts, in particular the infra-red screen and the two eye glass magnifiers. The screen should be obtainable from most large photographic suppliers but in case of difficulty some alternatives have been given in the text.

The two lenses used do not have to be exactly the same as those shown on the prototype but
they do give a neat finish to the unit. You should be able to obtain them from some watchmenders but you may have to hunt around.
Alarm requirements will decide if RLA 1 is to be used, the main point when buying this is that the contacts can switch the load applied by the alarm.
If you have an old OC 71 you could try scraping the black paint off it to use it as TR1 (OCP 71). The old types were filled with an almost clear jelly-like substance which allows light to get at the junction inside-newer ones use an opaque substance which does not allow enough light to pass.

## Capacitance Meter

When buying for the Capacitance Meter you must make sure that the capacitors you get are the correct type and that the tolerances are within those specified. You will probably have to pay a little more for the better tolerance but the resultant accuracy of the meter is worth the extra.
A two-pole six-way wafer switch has been used in the prototype although only five ways are used, this is because the six way type is easier to obtain.

Once again with this project the meter will be the most expensive part, but it is worth buying a fairly good one for the sake of accuracy. You should be able to buy one marked with $0-10$ and 0.3 scales.

It is of course, possible to recalibrate your own meter, using Letraset, if you carefully dismantle the case and remove the scale.

## LW/MW Radio Tuner

There should be no buying problems for the Radio Tuner, the only items that could possibly cause trouble are the two coils. If you cannot get them in your area write to one of the larger London based suppliers-most of them are able to supply.

## New Products

Having written about the Linear 505 amplifier last month, and complained about the poor specification quoted, we then received news of another new amplifier from Linear-the 606! Linear say that the 606 is believed to be the lowest priced stereo ampli-
fier, designed for a magnetic cartridge, to ever become available through the normal wholesaleretail channels. The recommended retail price is $£ 22 \cdot 50$.

The specifications quoted are rather better than for the 505: output, 6 W music power into $15 \Omega$ (that's about $3 W$ continuous r.m.s. [our estimate]); input sensitivity, 3.5 mV magnetic, 35 mV ceramic, 100 mV tape and 400 mV tuner; total harmonic distortion 0.1 per cent at 1 watt; frequency response range (whatever that means) 20 Hz to 65 kHz .

We said the specifications quoted are rather better, by this we mean the way they are quoted, not the actual figures. Since most of the figures are qualified in some way it is a pity that a proper frequency response figure is not quoted, but at the price, one must not expect the highest quality.

Another new product this month is the Mod-3 case from West Hyde Developments Ltd, Ryefield Crescent, Northwood Hills, Middx. HA6 1NN. West Hyde have long provided a professionally finished case (the Mod2) and they have now introduced this new case design.

Mod-3 cases are provided in a variety of sizes, the smallest being $7 \times 3 \times 5^{1}$ inches deep and the largest $11 \times 6 \times 5^{1}{ }_{4}$ inches deep, they are finished in blue and grey p.v.c. covered metal (outside blue, panels grey) and cost from $£ 2 \cdot 25$ to $£ 4 \cdot 25$.


## Bi-Pre-Pak Ltd.

Due to a printer's error in our August issue an incorrect price was quoted for the Complete Telephone as offered for sale by the above company. The correct price should have read 95 p. We offer our apologies to Bi-Pre-Pak Ltd., and to any readers who may have been inconvenienced by this error.

## FOR RAPID SERVICE <br> GARLAND BROS. LTD, DEPTFORD RROADWAY, LOWDON, SEE GQN

TRANSFORMERS
Miniature
MM6 $6 \mathrm{~V} .500 \mathrm{~mA}+6 \mathrm{~V} .500 \mathrm{~mA}$
$M M 66 \mathrm{~V} .500 \mathrm{~mA}+6 \mathrm{~V}, 500 \mathrm{~mA}$
$M M 1212 \mathrm{~V}, 250 \mathrm{~mA}+i 2 \mathrm{~V}, 250 \mathrm{~m}$ MM12 $12 \mathrm{~V}, 250 \mathrm{~mA}+12 \mathrm{~V}, 250 \mathrm{~mA}$
$\mathrm{MM20} 20 \mathrm{~V}, 150 \mathrm{~mA}+20 \mathrm{~V}, 150 \mathrm{~mA}$ $\mathbf{8 1} \cdot \mathbf{2 0}$ plus 13 p p. \& p
Tis.
$126.3 \mathrm{~V}, 3 \mathrm{~A}-80 \mathrm{p}$ plus $26 \mathrm{D}^{2} \mathrm{D}$ \& \& $p$
 LT4 12 V , $3 \mathrm{~A}-\mathrm{El} .32$ plus 30 p LTS 9-a-9V, 0.5A-75p plus $21 p$ p. LT6 12-0-12V, 1A-95p plus 26 p. LT7 30-0-30V, IA- 11.87 plus 30 p Multi-tapped
MT30/2 0-12-15-20-24-30V, 2aMT60/1 0-5-20-30 plus 30p p. \& p. MT60/2 $0-5-20-30$ plus 30 p p. 8 g Charger
62.90 plus 340 p. \& $p$ a


 Auto-transformers
AT30 30W-61-12 plus 30 po 8 D ATIS0 150 W - 2.55 plus 34 p p. \& op . AT300 $300 \mathrm{~W}=64-75$ plus 42 P p. \& aT1000 $1000 \mathrm{~N}-68.90$ plus 62 p All shrouded with terminal blocks AT30 0-110-240V. All others 0 -$10-200-220-240 \mathrm{~V}$.
Speaker isolating transformer
il ratio for $3-15 \Omega .2 W-860$ plu $13 p p$ \& \& $p$. Tapped 3, 8, $16 \Omega$. Will match almost any speakers to any amplifier
15 W max.-90p plus 20 p p. \& p .

## ALUMINIUM BOXES

with lids and serews
Type. L, Wricep. \&
GB7.
 GB9: in 27 in 1




 These sizes veroboards

EQUIPMENT CASES
with sloping front panel
Type $H$. $W$. D. Price $p$. $p$

 Stain aluminium. silver-grey ham-
mer finished, 20 p exera.

## CONSOLE CASES

in phain aluminium, ideal for mixers, Type W. A BC: D Price p. \& p. GB20 $\begin{array}{lllllll} & 9 & 3+2 & 3 & £ 1-42 & 30 p\end{array}$ $\begin{array}{lllllll}\text { GB21 } & 10 & 9 & 3 \leqslant 2 & 3 & £ 1.58 & 30 p \\ \text { GB22 } & 12 & 9 & 3 \frac{1}{2} 2 & 3 & E 1.72 & 30 \mathrm{p}\end{array}$


## VEROBOARD

## ELECTROLYTICS

| $1 \mu \mathrm{~F}$ | 450 V | $19 p$ | $1.000 \mu \mathrm{~F}$ | 25 V | P |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $2 \mu \mathrm{~F}$ | 500 V | 20p | 1,000 F | 50 V | 39p |
| $4 \mu \mathrm{~F}$ | 350 V | $14 p$ | 2,0 | 25 V | p |
| $8 \mu \mathrm{~F}$ | 450 V | 16p | 2,000 $\mu$ | SOV | 53p |
| $16 \mu \mathrm{~F}$ | 450 V | $17 p$ | 2.500 | 25 V | 45p |
| $25 \mu \mathrm{~F}$ | 25 V | $7 p$ | 2.500 \% | 50 V | 60p |
| $25 \mu \mathrm{~F}$ | 50 V | 89 | 3,000 | 25 V | 48p |
| 32, 4 F | 450 V | $24 p$ | 5.000 | 25 V | 55p |
| S0, F | 50 V | $10 p$ | $5.000 \mu$ | 50 V | 98 p |
| $100 \mu \mathrm{~F}$ | 25 V | 10p | 888 | 450 V | 18 p |
| $100 \mu \mathrm{~F}$ | 50 V | 10p | 8-16\% F | 450 V | $20 p$ |
| $250 \mu \mathrm{~F}$ | 25 V | 120 | ${ }^{16}$ - $16 ; 5$ | 450V | 27p |
| $250 \mu \mathrm{~F}$ | 50 V | 17p | ${ }^{16-324 F}$ | 450V | 63p |
| 500 ${ }^{\text {F }} \mathrm{F}$ | 25 V | 180 | 32-324 | 450V | 49p |
| $500 \mu \mathrm{~F}$ | 50V | 25p | 50-50, | 350 V | - |
| MINIATURE ELECTROLYTICS |  |  |  |  |  |
| $1 \mu \mathrm{~F}$ | 63 V | $6 p$ | 10y\% | 64 V | 7 p |
| 2-2, 5 | $63 \vee$ | Pr | 1614F | 40 V | 7 p |
| 4 AF | 40 V | $7{ }_{P}$ | 30, F | 15 V | 7 p |
| 47 cm | 63 V | $6 p$ | 474F | 16 V | 7 p |
| B. F | 15 V | \% | 47aF | 25 V | $6 p$ |
| 5 | 40 V | $7 p$ | 68uF | 16 V | 6 p |
| OLa ${ }^{\text {a }}$ | 25 | ${ }_{6 p}$ | 100․:F | 10V | 6p |
| ENTIRE MLILLARD 015016017 RANGE ALSO STOCKED |  |  |  |  |  |

## CASSETTE OWNERS!

For Philips and similar casserze recorders.
PU12 Power unir for connection to
syseems, giving $7 \frac{9}{9}$ v, seabilised $£ 3.25$ PUI4 As above but switched for $\mathbf{6 5 . 1 0}$ PP75 Main power supply.

E1.95 All units are complece with cable and olug. VARIABLE POWER SUPPLY
Input: 240 V, a.c.
Output: 5 witched $3,4.5 .6,7.5, ~$
O, 12 volts d.c. at 500 mA

## BATTERY ELIMINATORS

suitable for eransistor radios and similar lizht CuFrent equipment PP9 input 240 V a.c. c . Outpur 9 V d.e. Price $£ 1.50$ plus 12p D. \& p.

## NEW NEW <br> ILLUSTRATED 1972-73 CATALOGUE <br> Post Free

CONTROLS, Log. or Lin.
Single, less switch. 150
Single, DP switch. 240
Tandem. less switeh. 40
$5 \mathrm{~kg}, 10 \mathrm{k} \Omega, 25 \mathrm{kR}, 50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega, 250 \mathrm{k} \Omega$, $500 \mathrm{k} \Omega, 1 \mathrm{Mg}, 2 \mathrm{Ma}$

SLIDER CONTROLS 87 mm .
complete with knobs.
Single. 44 p ; Tandem, 55 p . $10 \mathrm{~kg}, 25 \mathrm{~kg}$. 50 kg . $100 \mathrm{k} \Omega$. log. or lin.

## RESISTORS

Carbon
All $5 \%$, hish-stability. EI2 values. IW, 1p: WW. Ipi iW. 4p; 2W. 6p
5W, $10 \mathrm{D} ; 10 \mathrm{~W}$, 12p

## LOUDSPEAKERS

Tin $\times 4 \mathrm{in}, 38-61 \cdot 12,8 \Omega-$ E1.12

10 in $\times 6$ in $32.3-62.32,8 \Omega-62.32$,
$158-62.32$.
8in round. 32 . $22.10,8 \Omega-62.50$,
Adastra"Hi-Ten", $10 \mathrm{in}, 10 \mathrm{~W}, 8$ or $15 \mathrm{~g}-\mathrm{E3}-40$. Please add 20 p p. ip. to all qpeakert.

## BONDED ACRYLIC FIBRE

B. A.F. wadding. IBin wide. lin thick. The
ideal lining for speaker eñelosures. 25 p per
友 yard.

## plugs

Cor aerial
D.I.N. 2 pin (speaker)
D.I.N. 3 pin D.I.N. 3 pin
D.I.N. 4 pin
D.I.N. 5 pin. 240

Jack, $2 t \mathrm{~mm}$ unscreene
Jack, $2 \frac{1}{2} \mathrm{~mm}$ sercened
Jack, 3 tmm unscreened
Jack, 3 mm screened jack, in unscreened
jack. tin screened lack, stereo. unsercened jack, sterco, sercened
Phono. plastic top Phono, plastic top Phono, plated metal
Phono, fited 4 ft lead Phono, ficted 4 ft lead
Wander, red or black Banana 4 mm . red or black

## LINE SOCKETS

## Car aerial

D.I.N. 2 pin (speaker) DIN. 3 pin
D.IN. 5 pin, $180^{\circ}$
D.I.N. 5 pin, $240^{\circ}$ Jack, $3 \frac{1}{2} \mathrm{~mm}$
Jack, $\frac{3}{2}$ minn
Jack, $\frac{1}{4}$ in screened Jack, stereo, screened

CAPACITORS

| ${ }^{50} 5$ |
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0.0022
$\begin{array}{llll}0.0022 \mu \mathrm{~F} & 500 \mathrm{~V} & \text { SiM } & 6 \mathrm{~F} \\ 0.0022 / \mathrm{FF} & 1,000 \mathrm{~V} & \mathrm{MDC} & 6 \mathrm{p}\end{array}$


## MAIL ORDERS: Some icems have a post and pecking <br>  

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Photogragh: Science Museum, London.

WE come now to our first British pioneer, but as he was not a physicist nor engaged in proving any electrical theory he may be termed "a stranger in our midst". James Watt inventor extraordinary; the man after whom the unit of electronic power is named (see Table 1).

The year is 1755; the place Greenock on the Clyde. Nineteen-year-old James Watt said goodbye to his family and set off for London, the hard way, by horseback to look for work. His father a small merchant had lost his trade and fortune through bad speculation. Because of this and ill health, Watt had been unable to go to school regularly and was therefore largely self taught.
Arriving in London some twelve days later, the young James obtained employment at the instrument works of John Morgan. After twelve months he returned home but was forbidden by the Glasgow City Guild to start a business as an instrument maker because he had not served a full apprenticeship.

## NEWCOMEN'S ENGINE

Watt obtained work with the college of Glasgow in a model making and repair shop. The college asked him to repair their model of Newcomen's engine which had been invented some sixty years earlier and had since been used to pump water out of coal mines.

Having got the model working Watt was amazed to find the great consumption of steam; he reasoned that he could produce a better version. In 1765 he made a large scale engine, which was erected at Kinneil near Linlithgow. This gave Watt the opportunity to go into the construction in more detail.

Table 1: WATT (W)


#### Abstract

The watt might be termed the "horse-power" unit of electronics. in fact 746 watts are equal to 1 h.p. The power needed to maintain a current of one ampere through a conductor, and a potential of one volt across its ends is equal to one watt. The unit was first proposed by C. W. Siemens in his presidential address to the British Association in 1889.


Large scale trials and patent fees took what little money he had, and Watt was forced to agree to Dr. John Rocbuck founder of the Carron Ironworks; taking two thirds of any profits from the invention in return for bearing costs, but the two partners did not get on well together and after a few years of uneasy collaboration they parted. Once again shortage of money prevented Watt from bringing his invention before the public.

## BOULTON AND WATT

In 1768 Watt met Matthew Boulton, a man of considerable vision who could see that steam engines need not be confined to pumping machines and that they had a great future. Boulton, a Birmingham manufacturer, was owner of one of the most modern engineering works in Great Britain. He agreed to take Roebuck's share in the invention and a new and famous partnership
was born. Then in 1769 , Watt obtained his first patent, his machine produced more power for its size than Newcomen's and used less fuel but was still only usable as a steam pump.
In 1781 he patented his second engine which converted the reciprocating motion of the piston rod into a rotary motion and "drove a wheel round". This opened up new frontiers and was the start of the real steam age; it set Britain on the road as a great manufacturing power.

The Watt family built Heathfield Hall, a mansion on Handsworth Heath, Staffs, on a forty acre site. Although by now wealthy and famous, Watt worked constantly in his garret workshop. His restless brain invented a sculpture copying machine, a machine for drawing in perspective, and a press for copying manuscripts.

Watt died at Heathfield on August 19, 1819.

James Watt, garret workshop

Photograph:, Crown copyright Science Museum, London.



## Knotted

A serious error was made by somebody when he knotted the mains input cable, illustrated twice (on Pages 530 and 531) in the article on the Drill Speed Controller.
This can be very dangerous and is specifically outlawed by the Institute of Electrical Engineers in their published regulations.
If a cable is rapidly bent from a straight to a highly twisted configuration, as often occurs when an item is knotted and the knot pulled tight rapidly, the core material may suffer from local fatigue causing cracking of the material. This occurs very readily with copper core, but less so with the more resistant steel ones. The cracking acts as a resistance to the current flowing through the appliances, hence local heating occurs. At the 200 watt ( 0.9 amps ) being drawn by a domestic drill this heating may be only a few degrees, but if the same knotted lead were used to feed say a three kilowatt fan heater the temperature may rise 100 degrees above ambient temperature, melting the cable insulation and possibly doing damage to someone or thing but at least blowing the fuse.
Hence although this may appear to be excusable in this instance I believe that it would be a good idea if a few words of caution were published in a future edition of your publication to "head off" anybody gaining the general impression that this is accepted safe practice which it most certainly is not.
The error, though dangerous, is so common that the writer could be excused it, provided the suggested warning is printed.
K. R. Kinsella, Aylesbury.

We thank you for pointing out our mistake, we would suggest that readers use one of the plastic cable clamps available, to secure mains leads.
For the Drill Speed Controller this could be screwed to the base of the MK box.

## Radio Amateürs

I write to ask that a brief item be inserted in Everyday Electronics magazine re. the amateur radio course run by Northumberland Education Dept. at Gosforth, very near Newcastle-upon-Tyne.
The course is designed to prepare students for the Radio Amateurs Examination in May/ June 1973. It will be run at the Gosforth Evening Institute, Gosforth Secondary School, Regent Avenue, Gosforth, Northumberland, commencing in September 1972.

Designed specifically for the R.A.E. the course is also ideal for anyone wanting to get an insight into radio theory, having just taken up radio or electronics generally as a hobby or professionally.
Held on Tuesday/Wednesday of each week from 7p.m. to 9p.m. candidates may sit the R.A.E. at the school.

Enquiries should be addressed to, The Principle, Gosforth Evening Institute at the above address who will forward a prospectus by return, or further information can be had from myself by telephoning Newcastle-upon-Tyne 668439.

As you will have gathered, I take the class and your co-operation in this matter would be greatly appreciated.
D. R. Loveday, Newcastle-upon-Tyne.

## Too Slow

I have just built the Electronome from the circuit as described in the July issue of Everyday Electronics. I have used all the correct component values, and have checked all wiring. You say that this circuitshould give 40-225 beats per minute.I only get 18-100 beats per minute.

Could you please advise me what may have gone wrong.
N. Matheson,

Newcastle-upon-Tyne.
The reason for your Electro-
nome not giving the correct range of beats per minute is almost certainly that the capacitors are too high in value. This is quite common as electrolytic capacitors have very poor tolerances, e.g. $-20 \%$ to $+100 \%$ of the nominal value. We suggest that you reduce the value of C 2 .

## Soldering Irons

Could you please help me in buying a soldering iron? I am a beginner toelectronics and before the end of the year the firm I work for will be going over to electronic calculators from mechanical types.
I have managed to get all the issues of Everyday Electronics to date in order to get myself used to electronics; I have a Demo Deck and plan to try some of your projects, but I can't make my mind up which would be the best soldering iron for the projects in your books and for future use on other projects.
Could you please explain the connection between the iron's leakage current to the electronic components, and the damage that can occur.

Finally, the bits for the iron are available ""nickel plated," "Ferra* clad" and "Triple coated," which do you use for what job?

I would like to thank you on behalf of us beginners for bringing out a magazine which we can understand. I only hope that you don't take too much notice of some of the letters that you have published from people who want you to turn into the same type of magazine as all the others, leaving us beginners once again on the outside.

There is always a new generation of beginners coming along who need a magazine like this as a stepping stone before going to the more experienced magazines with all the technical jargon.

> G. Hayes,

Hackney, London.
Provided you are soldering to a circuit that is not connected to a supply or to earth no damage can result from iron leakage. When the circuit is live and part of it is earthed or connected, via a transformer to the mains supply a low leakage iron should be used to avoid possible damage to semiconductors.

The more expensive coated bits are made to last longerthey are designed not to corrode as much as a normal copper bit. 'Coated bits are generally used on assembly lines but they are good if you tend to leave the iron


MAINS OPERATED CONTACTOR 220/240v. 60 cycle solenoid Tlth Imminited core so very alient in operation. closes circuits each ratid si 10 ampe. Gertann Electrical Compy Overall size $21 \times 2 \times 2$. E1. 50 each.


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tablc. Buftable for inv posilive or negative earth 8 opplied complet with fitting instructions and with gitting instructions and reads
wired dialhboard switch. $E 5.75$ plar 25p poot and insurance.

## MAINS TRANSISTOR POWE

 Dezlgned to oper PACKDeslgned to operate transistor scts and amplisers. Edjustable outpal 8 v ., 9v, 12 voite for gp to of the following batteries: PPI, PRS, PP4, PPB PP7. PP9 and others. Fit comprises: masins traneformer reetiner, stooothing and load resistor. condensers and instructions. Real anip at onlr
f1, plos 1 \&p postace.

## MICRO SWITCH




MINIATURE
WAFER SWITCHES
$\frac{2}{2}$ pole, 2 way-t polc. 2 waypole, + was-3 pole, 4 way -2 pole
6 way -1 polc, 12 wy . All at 200 Exay- 1 polc, 12 way. All at 200
KITS FOR PREVIOUS PROJECTS Unlcks otherwise stater. kits contain clec-
tronio parta sooty. The cace and apecial tronic parta ools. The case and apecial
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lemnl oemlator coil 75p cach. 10 for 86.75 . 100 for $262-50 \mathrm{p}$. Witb connection dig.

|  | Standard size $1{ }^{\prime \prime}$ wafer-silver-piated 5 -2mp contact standard $i^{\prime \prime}$ apindle $2^{*}$ long-with locitng washer and nut. |  |  |  |  |  |  |  |  |
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| No. of Poles | 2 w 25 |  |  |  |  |  |  |  |  |
| 1 pole | 40p | 403 | 40D | 40p | 400 | 40p | 400 | 40 p | 0 |
| 2 poles | 40 p | 40 p | 40 p |  | 40p | 40 p | 40p | 700 | 70p |
| 3 poles | ${ }^{40 \mathrm{p}}$ | 400 |  |  | 709 | 70p | 70 | ${ }^{95 p}$ | ${ }^{95 p}$ |
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| 5 poles | 40 p | 400 | 700 | 7 | ${ }^{95 \%}$ | ${ }^{95 p}$ | 950 | 21.45 | E1.45 |
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|  | ${ }^{700}$ | 700 | 95 p | ${ }^{25}$ | 21.45 | 81.45 | 21.45 | 122.4 | 28.45 |
| 10 polea | 70 p | 70p | 95p | 21.20 | \$1.45 | 21-45 | 21.45 | \$2.70 | ¢2.70 |
| 11 poles | 70p | 950 | 85D | E1.20 | E1-70 | \$1\%0 | E1.70 | 22.95 |  |
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cased in an lvory plantic case with clear plastic windown, cared in an lvory plantic case with clear plastic windowne,
themoneter above and swltch setting scalc below. Slze approx. $\$-8^{\circ} \times 3 \cdot 2^{-} \times 1 \cdot 4^{\prime}$ deep. Can be mounted on approx. 3 conduit box or directly on wall. Price 21.5 each or 10 for conduit
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| Mr91s | is | AA129 | 15p | BAY18 | 17\% | 0 OS | 17 |
| IN916 | 7 | AAZ13 | 12p | BAY31 | 7 | O410 | 20D |
| ISs $0^{\circ} \mathrm{O}$ | 90p | AAZIS | 12p | BAY38 | ${ }^{25 p}$ | OA9 | 10p |
| IS:4 | \% | AAZ17 | 10p | BY100 | 159 | OA47 | 8p |
| 15113 | 15p | BA100 | 15p | BF103 | 29 | OA70 | 7 7 |
| 18120 | 120 | BA102 | 25 | BYI22 | 47ty | 0873 | 100 |
| 18121 | 14p | BA110 | 25p | BY124 | 15p | OA79 | $7{ }^{7}$ |
| ISISO | 80 | BA114 | 15p | BY126 | 15 D | OA81 | 85 |
| IS131 | 10p | BAl15 | 7 P | BY127 | 170 | OA85 | 100 |
| 18132 | $12 p$ | B4141 | 17p | RY16s | 57p | OA90 | 7 p |
| 18920 | $7{ }^{\circ}$ | BA142 | 17p | BYX10 | 29 | OA91 | 7 7 |
| 1S922 | 8 p | BA144 | $12 p$ | BYZ10 | 35 | OAPS | 7 |
| 18923 | 12p | Bal4s | 17p | BYZ11 | 320 | OA200 | 70. |
| L9940 | 5p | BAX13 5p |  | BYZ13 250 |  | TIV307 | 10p |
|  |  |  |  | 50p |  |
| "SCORPIO" CAP <br> DISCHARGE IGNITION SYSTEM <br> (As printed in P.E. Nov. <br> '7I). Complete kit $\mathbf{£ 1 0 - 0 0}$ $\text { P. R P. } 50 \text { p. }$ |  |  |  |  |  | BRIDGE RECTIFIERS |  |  |  |
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|  |  |  |  | 1100 | 37 p | 450 |  |
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| $23 \times 5 \mathrm{in}$ | 25 p | 25 p |
| $31 \times 3$ in | \%5 | 25 p |
| 3t $\times \sin$ | 30 p | 29D |

$5 \times$ 17lin (Plain) 83 p,
Vero Pins (Bac of 36) 20
Vero Pins (Bact of 36) 20p
Vero Catter 45p
Fin Intertion
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| OPTOELECTRONICS |
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| TK 2O9 LGETS ENTTTIMG |
| DIODE (RED) 35P. |
| B990 PHOTORESISTOR 38p |

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1.5 F
$1 \cdot 2 \cdot \mu \mathrm{~F}$
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WIRE-WOUAD RESISTORS 2.5 wratt $6 \%$ (UP to 270 obras 5 oniy). 7 p (up to $6.2 \mathrm{k} \Omega$ only), 9 10 Fatt $5 \%$ (up to $25 \mathrm{k} \Omega$ only).

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0.1 Watt $6 p$ VERTICA $\begin{array}{llll}0.2 & \text { Watt } & 6 p & \text { OR } \\ 0.2 & \text { Watt } & 71 p & \text { HORIZONTAL }\end{array}$

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PA1077 80p
 stock. Please enquire.
on for long periods without making any joints. Iron manufacturers will be able to provide you with more information on particular types of bits.

## Bug Report

Four months ago you published some of my reflections on how I was feeling as a recently infected victim of the "Electronics Bug", so perhaps you may be interested to hear something of the post-infection period.

At the present time temperature is now back to normal, or at least I don't get quite so hot under the collar trying to follow some of the more complicated theoretical items, such as how an actual circuit performs. You see I have now resigned myself to the fact that it is perhaps a little late in the day to really learn thoroughly how and why it all happens.

With only a limited amount of time to devote to a hobby which is after all to some a very fulltime occupation; I have decided that maximum pleasure will be obtained if I cease puzzling how a particular circuit functions, and just get on with, what is for me the most enjoyable part of the activity, the actual construction of a project.

No doubt some of the more erudite purists among your readers and contributors will throw up their hands in horror at this "short cut" attitude. My defence is that each of us must know best how much time and effort we can afford to devote to a hobby, and the criterion is surely the amount of pleasurable relaxation that can be derived from it.

In my own case the main pleasure is the construction of the housing of radio receivers, I find a great deal of satisfaction out of designing a suitable cabinet, in the main using plywood either varnished, or covered with self-adhesive plastic of which a huge variety can be very easily obtained almost anywhere. The latter method does of course have the advantage of speed, as you don't have to wait between the several coats of varnish that are necessary to obtain a really pleasing finish.
For the Astron (which incidentally really did "work first time" as you forecast!) I didn't consider I could make a good job of the perspex case featured in your article, also I was anxious to get en with it and so looked around for something quickly available. It was felt necessary to keep to the spirit of the project, I finally selected a transparent rigid plastic lunch box and managed to fit
everything neatly with only very minor alterations to the suggested lay-out.

There are several sizes of rigid plastic containers available in most hardware and chain stores, usually intended for food storage, picnics, etc., and are worth considering when one is searching for something quick and easy.

Naturally, for those who require something even easier there are plenty of diecast boxes and instrument cases to be had in all shapes and sizes from many of the component suppliers who advertise regularly in your columns.

As this progress report has become more of a case history (ugh!) perhaps I could continue by referring to the excellent article in March E.E. namely Cases from Chassis which appeared just as I was contemplating construction of the Electro Laugh so this seemed an ideal time to put the method into operation.

I sent off my $£ 2$ to the supplier who regularly offers a kit of parts for your projects and was delighted to find the order was dealt with very promptly indeed. I was especially pleased to find that not only was the approximate cost quoted in your article an apparently realistic one but that the kit also contained the extra parts to include the optional blanking gate. Full marks all round!

All the components and circuit board were fixed directly to a panel (you've guessed it-plastic covered ply-wood!) cut to fit the open side of an 8 inch $x 6$ inch aluminium chassis all as per the Cases from Chassis article.

With perseverence, I am sure the smaller size chassis could have been utilised, but as size was not an important factor this
time I found it simpler to use the larger case which gave more scope for planning a pleasing layout, bearing in mind this project uses two 6 volt batteries.

The "speaker" supplied with the kit, a $2^{1}{ }_{2}$-inch diameter telephone ear-piece was found to give excellent volume, much better than various impedance mini-speakers I experimented with, so this was mounted centrally on the front panel, the moulded flange on the ear-piece making fixing a simple matter of cutting a circular hole in the panel, and applying a little contact adhesive.

The blanking gate was wired via a second push switch, so the type of "laugh" produced depends on whether one presses one, or two buttons, and can of course be varied as the mood takes you.

In actual fact the sound produced is not really very much like laughter, or at least it isn't on my model. It is more like a cross between a wailing banshee, and a soul in torment, with police siren overtones. However it certainly makes me laugh, also those friends and colleagues who are still speaking to me after exposure to its somewhat strident tones.

To sum up, an interesting unusual project, of no real practical value other than a (end of!) conversation piece. As a laughter simulator: 5 out of 10 ; but as a laughter stimulator-full marks!
J. G. Richards

Sale.

We must point out that we are unable to supply the back issues mentioned in this letter. Nor are we able to supply copies of individual circuits or articles.


The approximate cost of components for the Weather Station (featured last month) should have been given as $\mathbf{5 7} .50$, excluding the two cases but including the vanes and rotor kit. Also in this article D1 is a $3.3 \mathrm{~V}, 400 \mathrm{~mW}$ Zener diode.

Drill Speed Controller (featured last month) see page 614.

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$240 \mathrm{ohm}-50 \mathrm{p}$ each
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o. 2.5 mm (please state which type $20 p$ plus $3 p$ p. $* p$. BMINAPREFRR BARGAIE EM.I. 450 set 2. 8. 15
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$000,1,000 \mathrm{~V}$ ． D．C．Current： $10,100 \mu \mathrm{~A}$ ， $10,100,500 \mathrm{~mA} .2-5,10$ amp．Reaistanc $1 \mathrm{~K} .10 \mathrm{~K}, 100 \mathrm{~K}, 10 \mathrm{MEG}$ $100 \mathrm{MFG} \Omega$ ．Decibels：-10 to +49 db ．Plas－


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 \begin{tabular}{cc|cc|c|c|}
CA3005 117p \& FJE151 \& $25 p$ \& SN7442 \& $75 p$ <br>
CA3007 \& $262 p$ \& FJH161 \& $70 p$ \& $8 N 7446$ \& $100 p$

 

CA3011 \& $75 p$ \& FJHITI \& $25 p$ \& ENT447 \& 185p <br>
CAS012 \& 880 \& PJH181 \& $25 p$ \& $\$ N 7448$ \& $125 p$

 GA3013 105p FJH281 250 SN7450 20p 

CASO14 124p \& FJH231 \& $25 p$ \& $8 N 7451$ \& $20 p$
\end{tabular} CA 3019 CA3020A

CA3091
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| OA2 | 35 p 12 | 2584 | 3001 5 | EM80 | ${ }^{45}$ |
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| OB2 | 459 | 2575 | 429 E3 | 3r81 | 60 |
| OZ4 | 30025 | 2526 | 659 | Y884 | 85p |
| 144 | 20080 | 30 Cl 5 | 80 p | EM85 |  |
| IR5 | 40 p 30 | 30 C 17 | 90 p E | M87 | 700 |
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