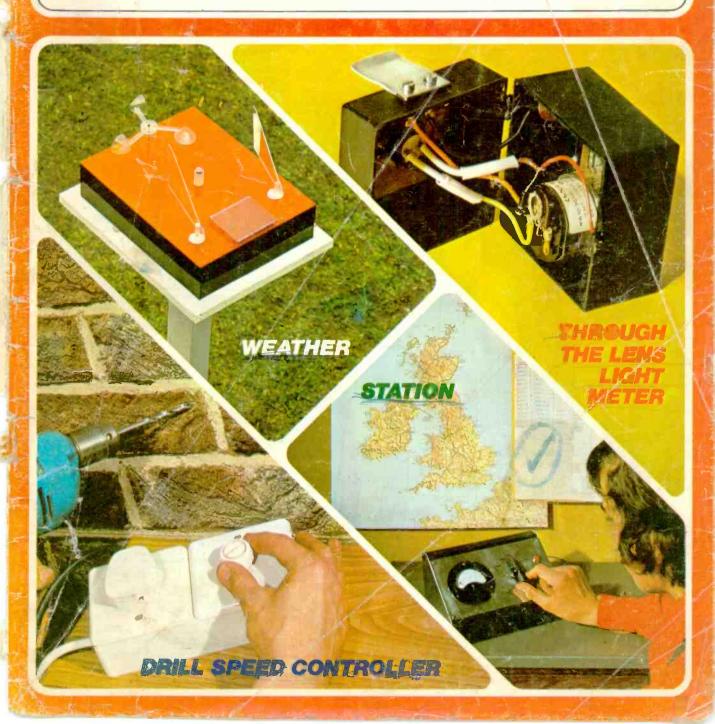
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2N706			17p 2N3704						
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0 - 037; 0 - 056; each 3p.
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0 - 018; 0 - 022, ea. 5p; 0 - 22; 0 - 33, 6p; 0 - 39 7p;
0 - 018; 0 - 22, ea. 5p; 0 - 22; 0 - 33, 6p; 0 - 39 7p;
NIEMENS ELECTROLYTIC
with axial leads: Values in µFIV
0 - 47/100; 1/100; 2 - 2/63; 4 - 7/35; 10/25; 22/16;
20/163; 22/155; 47/35; 100/16; 100/25; 220/6, 220/10;
20/163; 22/155; 47/35; 100/16; 100/25; 220/6, 220/10;
20/163; 470/35; 100/16; 100/25; 220/6, 220/10;
20/163; 470/35; 100/16; ea. 14p;
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1000/25; 16e; 470/63; 1000/35; 100/10, ea. 10p;
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4/10; 4/40; 5/64; 6-4/6-4; 6-4/125; 8/4; 8/40; 10/2-5; 5/64/4
4/10; 4/40; 5/64; 6-4/6-4; 6-4/125; 8/4; 8/40; 10/2-5; 5/64/4
4/10; 4/40; 5/64; 6-4/6-4; 6-4/125; 8/4; 8/40; 10/2-5; 5/64/4
4/10; 4/40; 5/64; 6-4/6-4; 6-4/125; 8/4; 8/40; 10/2-5; 5/64/4
4/10; 4/40; 5/64; 6-4/6-4; 6-4/125; 8/4; 8/40; 10/2-5; 5/64/4
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				avallable	(see	note belo	w)
c	1/20W	5 %	82 Ω-220K Ω	E12	9	8 - 0	7
č	I/BW	5 %	4 - 7 Ω-470Κ Ω	E24	ī	0.8	0.7
č ·	1/4W	10%	4 · 7 Ω-10M Ω	E12	1	0.8	0.7
č	1/2W	5 %	4 · 7 Ω=10M Ω	E24	1 - 2	1	0.9
č	IW	10%	4 · 7 Ω-10M Ω	E12	2.5	2	1 - 8
MO	1/2W	2 %	10 Ω-IM Ω	E24	4	3	2 nett
ww	IW	$10\% \pm 1/20 \Omega$	0-22 Ω-3-9 Ω	E12	7	7	6
ww	3 W	9/	Ι Ω-ΙΟΚ Ω	E12	7	7	6
ww	7W	5 %	Ι Ω-10Κ Ω	EI2	9	9	8

Values: E12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56.
E12 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36, order.)
43, 51, 62, 75, 91 and their decades.

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	available	(see	note belo	w)	
(Ω)	E12	9	8 - 0	7	
(Ω)	E24	1	0.8	0.7	
Ω	EI2	1	0-8	0.7	
Ω	E24	1 - 2	1	0.9	
Ω	E12	2.5	2	1.8	
Ö	E24	4	3	2 nett	
Ω	E12	7	7	6	
Ω	E12	7	7	6	
Ö	EI2	9	9	8	

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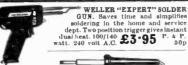


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OC4I	0 - 15	AUYIO	1 - 25
OC44	0.13	25034	0 - 25
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OC7I	0-10		
OC72	0.10		0.10
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254F	25V	7 p	2,500µF	SOV	60p
25µF	50V	.8p	3,000µF	25V	48p
324F	450 V	24p	5,000µF	25 V	55p
50 µF	50V	10p	5,000µF	50V	98p
100µF	25V	10p	8-8µF	450V	18p
100µF	50V	10p	8-16µF	450 V	20p
250µF	25V	12p	16-16µF	450 V	27p
250µF	50V	17p	16-32µF	450V	63p
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4.74F	63V	6p	47µF	16V	7 p
BuF	15V	7p	47µF	25V	6р
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jack, 3½mm unscreened
jack, 3½mm screened
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Jack, ½in screened
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125V

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S/M Disc P.S. Poly. M.F. Poly. Cer. S/M MDC MDC · 01 μF · 01 μF · 01 μF · 01 μF 125 V 160 V 250 V 400 V 500 V 10 p 7+p 0.01 µF 500V 0-014F 600 V 0 . 0 LuF 1,000 V

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everyday electronics

PROJECTS...
THEORY.....

CONVERSATION PIECE

We talk about it a lot. Usually to complain, more rarely to praise. Countless conversations have been sustained on this single topic alone. Weather.

All the paraphernalia of modern science and technology is harnessed in the cause of weather forecasting. Earth satellites report by radio what their electronic instruments detect high up in the atmosphere and computers devour this data and then produce their predictions.

From such sophisticated methods, let us now come back to earth and see what can be done in the ordinary garden or backyard, with some quite simple electronics. The *Weather Station* featured this month is not a scientific instrument of known and precise accuracy. Neither is it, on the other hand, simply a toy. With its convenient remote monitoring unit, the study of the weather can become a new absorbing interest.

THE MEANS

The means, no less than the end, are of interest, too. The *Weather Station* employs sensors and transducers that convert physical phenomena such as light, heat, moisture, and wind speed and direction into electric currents. These electric signals are used to produce a visual indication on a meter.

Employed here in this simple system are the same basic principles that are exploited on a far grander scale in all manner of advanced electronic equipments used for industrial and professional purposes.

The Weather Station might well provide valuable meteorological data to back-up that casual conversation in pub or train. It will certainly give the student of electronics a convincing demonstration of the inter-action between natural forces and certain electronic devices. Properties that provide the key to much modern technology.

RESIST!

At least one reader has noted that for the resistor, EVERYDAY ELECTRONICS uses a circuit symbol different to that recommended as first choice by the British Standards Institute.

We are not by nature rebels, but in this instant we do feel that the long established British zig-zag is, by far, a superior representation of the resistor (and its properties) than that nondescript oblong box, long favoured by the Continentals and more recently imported into the U.K.

If the pint is sacrosanct, and Parliament decrees that it shall remain despite impending national conversion to metric measurement, then surely the same privilege should be accorded to the traditional British resistor symbol!

fred Bennett

Our September issue will be published on Friday, August 18

EDITOR F. E. BENNETT • M. KENWARD • B. W. TERRELL B.Sc.

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.. EASY TO CONSTRUCT .. SIMPLY EXPLAINED



VOL. 1 NO. 10

AUGUST 1972

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This Drill Speed Controller is an extremely useful addition to the handyman's workshop. The unit will provide continuous control of speed of any a.c. mains operated series wound brush motor—most commonly used in hand electric drills and food mixers. Speed control is from approximately half full speed (which could be in excess of 1,000 r.p.m.) down to less than 15 r.p.m. The advantage of using electronic control is that one can obtain this dramatic variation without necessarily losing torque.

Even the newcomer to electronics will probably be surprised by the absolute simplicity of the circuit to be described (only five "electronic" components are used) and the unit can be made in one evening for a fraction of the cost of a commercially made equivalent. Making use of a standard electrical wall box simplifies mechanical construction and gives a very professional looking end product.

THE THYRISTOR

A thyristor or controlled silicon rectifier (CSR) is used to control the main current supplied to the motor; as this may be a new component to some readers we shall briefly describe the functions of this type of device.

Its symbol and designation are shown, together with a typical type in Fig. 1. The similarity of its symbol to that of a diode is deliberate because, to some extent, it shows similar properties. The anode and cathode are the main electrodes through which current can be made to pass and the third electrode—the gate—is used to effect control of this current.

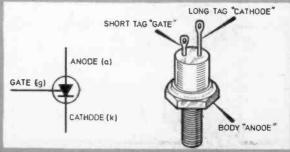
If connected into a circuit passing alternating current the thyristor presents a high resistance when the cathode is positive with respect to the

anode (just like a conventional diode) but unlike a conventional diode it also presents high resistance the other way. If, however, the anode is made positive with respect to the cathode and at the same time the gate is made positive with respect to the cathode the device will pass current in the forward biased direction. The reverse characteristic is not affected by the gate.

The interesting thing about a thyristor is that once it starts to pass forward current it will continue to do so even though the positive signal on the gate is removed; the only way the device can be "turned off" is for the current passing through it to fall below a certain value—this is called the holding current.

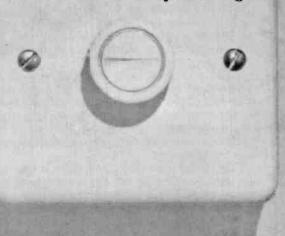
If we were using a direct current supply we would somehow have to interrupt the flow of current; in the case of a.c., however, the direction of current flow is reversed during every cycle of the mains and this effectively breaks the flow and the device will turn off The thyristor then stays off until such time that the gate is once more made positive with respect to the cathode.

Fig. 1. Circuit symbol and typical controlled silicon rectifier.



Electronic speed control for small electric drills and food mixers.

By Mike Hughes



Approximate cost of components £2:10 inclusive

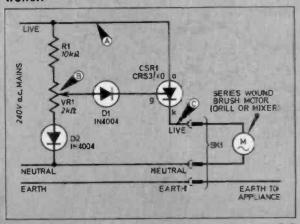
THYRISTOR TYPE

Unlike a transistor the thyristor has no linear slope to its characteristic and is either presenting very high or very low resistance to the flow of current. This means that there is never a great deal of power dissipated within the device and its temperature does not rise excessively even though it may be handling high currents at high voltages.

The thyristor we shall use (CRS3/40) has reverse and forward blocking voltages of 400 volts and when switched on can handle currents up to 3 amps. To switch on we must make the gate at least 2 volts more positive than the

gate at least 2 volts more positive than the cathode and allow at least 2) milliamps gate current to flow for about 1.5 microseconds (this is approximately the time it takes the device to switch on and for holding current to be

Fig. 2. Circuit diagram of the Drill Speed Controller.



established). The holding current minimum value is approximately 10mA—if the anode/cathode current falls below this level the device turns off.

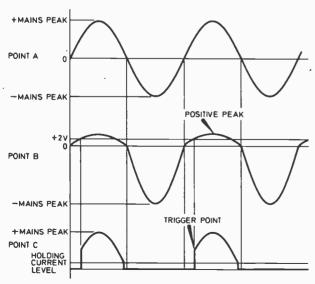
CIRCUIT DESCRIPTION

The circuit of the speed controller is shown in Fig. 2. The simplicity of the circuit is a little deceiving because the way it works is not quite as obvious as you might think at first sight. As we shall be operating from alternating current we shall assume that all voltages and waveforms are described relative to the "neutral" line, i.e., the line feeding the cathode of D2 and the motor. The "live" line will thus alternate from positive to negative.

During negative half cycles CSR1 will be reverse biased and can never be made to conduct therefore no power will be fed to the motor. For identical reasons D2 will not pass current therefore there is no dissipation within R1 and VR1.

During the first positive half cycle the potential difference across CSR1 will build up in the forward biased direction but as yet no conduction will take place. However, while this is happening D2 becomes forward biased and the potential at the wiper of VR1 (point B) will be rising in a positive direction—the peak voltage it can rise to is set by the potential divider effect of the wiper's position. As no current is, as yet, flowing through the motor the potential at the cathode of the thyristor will be the same as at the cathode of D2, thus we can say that the

Fig. 3. In the absence of feedback voltage, these graphs show the voltage variation at points A, B and C. The points in time over which the trigger point can be controlled are from zero to halfway through the positive half cycle.



potential at point B is rising positive with respect to the cathode and D1 becomes forward biased.

As soon as the potential at point B rises to about 2 volts positive with respect to the cathode of CSRI (assuming the wiper has been set at a position where this can happen) the thyristor will start to conduct and the full mains voltage appears across the motor.

Diode D1 now becomes reverse biased and no more gate current flows but because the thyristor is passing much more than the holding current it does not turn off. All this happens before the mains positive half cycle has reached its peak.

As the positive half cycle returns towards zero the current through the motor will fall—eventually to below the holding current—and the thyristor switches off; it stays off during the negative half cycle and in the absence of any other effects would be triggered on again some time during the first half of the next positive half cycle.

FEEDBACK

By adjustment of VR1 one can alter the position in time of the instant point B reaches 2 volts relative to the time of the start of the positive half cycle; thus we can control the length of time before the thyristor is triggered and mains is applied to the motor during the early part—the first 90 degrees—of the cycle. If we trigger early in the cycle more total power will be fed to the motor than if we triggered later.

Because, at best, only the positive half cycles can be used to power the motor it will not run at its full speed because we are halving the total energy supplied. By adjusting the total energy fed to the motor by setting VR1 at different levels we can influence the speed but at the expense of torque.

This circuit however goes quite a bit further and compensates for the apparent loss of torque at low speeds. This is brought about by a signal that is fed back to our circuit from the drill itself.

There is always a certain amount of remnant magnetism left in the iron that makes up the field coil of a motor and if the armature is spun within this remnant field the motor will behave like a dynamo producing an output voltage—in a series wound brush motor this will be in the form of a d.c. level and will be directly proportional to the speed at which the armature is turning.

You can see this effect quite easily if you connect a 10V d.c. meter across the input leads to the motor (it must not of course be plugged into the mains while doing this) and spin the chuck—if it is a drill—by hand. Remember while doing this to press the starting switch otherwise no current can flow through the meter! Even at low speeds you can detect several volts.

Components....

SHOP TALK

Resistor

R1 $10k\Omega$

5W wirewound

Potentiometer

VR1 $2k\Omega$

2W wirewound

Semiconductors

CSR1 CRS3/40 or any 400V 3A thyristor

D1 IN4004 D2 IN4004

Miscellaneous

3 inch length of 2 inch wide tag strip, small aluminium bracket—for mounting CSR1, insulated knob—preferably push-on type, MK double wall box type 2025, MK 13 amp unswitched socket, MK standard cover plate, three core mains cable—length as required, two 4BA countersunk screws—to mount circuit board, short length of connecting wire, grommet, mains plug—fused 13 or 15A type.

TORQUE

If we assume that this can happen to the motor in our circuit a rather interesting state of affairs is set up. During the negative half cycle following the first cycle when we set the motor going the armature will freewheel because of its inertia and now instead of the potential at the cathode of the thyristor being the same as that of the cathode of D2 (as was the case before switch on) the potential at point C will be a few volts positive—depending on the actual speed of rotation of the armature during its freewheeling action.

This voltage will be maintained well into the next positive half cycle. This means that the potential at B will have to rise to a higher level if the gate is to be made 2V more positive than

the cathode; thus firing during this positive cycle will occur later (it might not even occur at all if the wiper of VR1 is set low down). This reduces the energy fed to the motor and it will slow down until the next cycle when the "back voltage" produced will be less and triggering will occur slightly earlier in the cycle.

If the motor is subjected to heavy torque there will be a tendency for the motor to slow up dramatically during the freewheeling (negative) half cycle and thus the feedback voltage will be less allowing extra energy to be fed in when the next trigger cycle comes along. This extra energy will compensate for the increase in torque.

The net effect of this is that the motor will keep running at its lower speed but the amount of energy being fed to it will be hunting up and down as torque is applied and removed. The speed is altered simply by setting the potential at which point B can fire the thyristor.

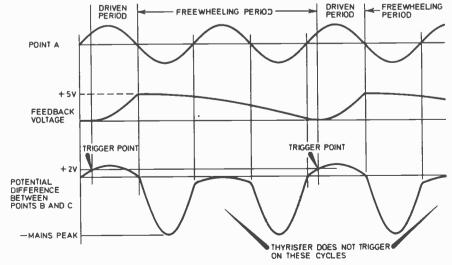
When running at very low speeds with very little torque applied the freewheeling action of the armature may continue over several full cycles before extra energy is applied. The motor will appear to "hunt." The effect is known as "skip cycling" and is inherent to this sort of circuit; as soon as torque is applied the skip cycling will stop and very steady low speed operation will take place.

The waveforms at points A, B and C are shown in Figs. 3 and 4, these may help in explaining the method of operation.

CONSTRUCTION

All components are mounted on a 3 inch length of 2 inch wide tag board which will ultimately be mounted in one half of an MK double wall box. The advantage of using this box is that not only does it make a very convenient and nice looking housing but is amply insulated because you must remember that all the

Fig. 4. A typical example of low torque allowing the feedback voltage to suppress triggering for the whole of one cycle (skip cycling).



DRILL SPEED CONTROLLER

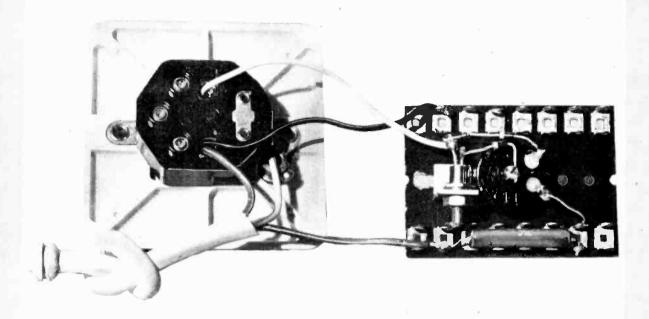
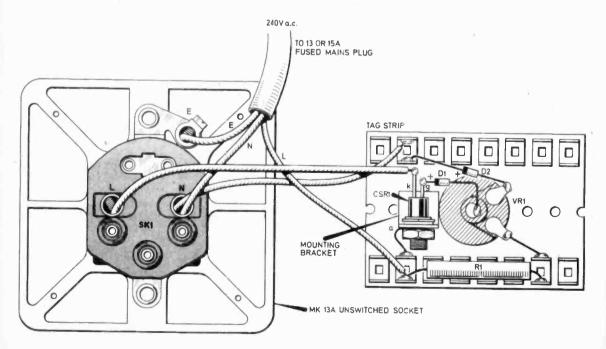


Fig. 5. Complete layout and wiring of the Drill Speed Controller.



circuitry is live to mains voltages.

The thyristor is mounted on a small aluminium bracket that is bolted to the tag board. The bracket acts as a small heat sink and easily disposes of any heat generated by the thyristor. Resistor R1 must be a 5 watt wirewound device and VR1 should have a rating of at least 2 watts and again should be wirewound. There is quite some latitude in the ohmic value of the latter.

If you cannot get a 2 kilohm potentiometer any value up to 5 kilohm will do but the range of control will tend to be cramped towards one end of the movement. The diodes D1 and D2 must be capable of withstanding full peak mains voltage and although the current passed is comparatively small the most convenient devices are 1N4004 one amp rectifiers, which are easily obtainable and relatively cheap.

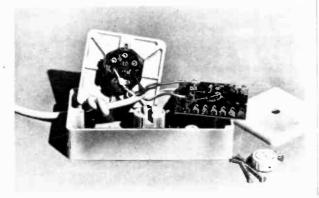
Mains input to the unit is from a 13 or 15 amp fused plug connected to a flying lead entering the case through a grommet. The output to the motor is via a standard MK 13A socket that conveniently bolts into the other half of the box.

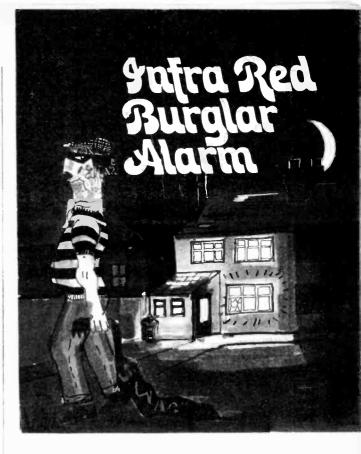
Once the component board is completed and wired to the socket (Fig. 5) it should be fixed—using countersunk screws—to two of the recessed tapped holes of the box and a standard cover plate is drilled to allow the spindle of the potentiometer to protrude. Screw the cover plate into position using the other two tapped holes and fix an insulated knob on to the shaft. Make sure that no metal parts of the potentiometer can be touched just in case there is an internal short.

OPERATION

When using the controller with an electric drill it is just as well to remember that, although the drill may be turning at low speed, if the torque applied is high you might be applying more electrical energy than you think and as some drills have a cooling fan coupled with the armature there could be an increase in temperature of the drill itself.

Provided common sense is used this does no more harm than the occasional overload but if you allow overheating to persist without sufficient cooling time you could burn out the armature!





Protect your home from any unwanted intruders with the invisible beam of our burglar alarm. As soon as the infra-red beam is broken a warning is given for 1 minute, the alarm then reverts to standby.

Capacitance Meter

Sort out any unknown or unmarked capacitors with values up to $3\mu F$ with this meter. Simple to use the meter will form an excellent piece of test gear for the work bench.

LW MW Radio Tuner

A simple long and medium waveband tuner for use with any amplifier, tape recorder or record player.

All in the September Sesue of



On sale Friday August 18.



Do you envy those fortunate folk who own the last word in test or hi-fi equipment? Do you sometimes wish that you could enjoy the same standard of professionalism in your home? If only it was not so expensive. Well now you can; your equipment may not be to the very highest standards, but price for price it will be of a higher standard than ordinary equipment.

You can do this by taking advantage of the many types of construction kits for radio, hi-fi and electronic equipment that are advertised by many firms and, more particularly, for hi-fi, the v.h.f. radio tuner and record player kits available in many types, prices and sizes.

CONSTRUCTION

It is not necessary for you to have a degree in electronic engineering, or that you should be a specialist in electrical equipment. Many of the kits now on sale are easy to build, both in time and effort, and will provide a great deal of pleasure, both in the building, and in their use afterwards.

Why build a kit? Here are two reasons: firstly, there is a saving in cash, exactly how much will depend on the piece of equipment you intend to build.

If you were to build, say, a high quality record

player, while it would not necessarily be cheaper than a commercial record player, it would certainly have a better performance than a comparably priced commercial unit.

Secondly, you could of course build to your own design, but in the first place this assumes that you have the knowledge to design such a unit and, secondly, although you might well end up with a sophisticated design, the finished unit would probably have a distinctly "home made" appearance. Kits do have the considerable advantage of professionally styled appearance, which would otherwise be difficult to achieve at home.

Quite apart from the knowledge that you will have created something that will be acceptable to your family and friends, when the time comes, who would be better qualified to service it than the man (or woman) who built it.

TYPES

Basically kits come in two types: The "little bits" type, in which, when you open the box,

The heading photograph shows a Transona Five radio kit from Radio Exchange Co., under construction. This type of kit is suitable for a first project.

you find yourself with all the individual components, one at a time, and the obvious assumption that you have enough electrical knowledge to sort them all out for yourself.

Then there are the "big bits" type where, on opening the box, you find yourself the proud possessor of a number of ready wired panels, or modules, which need only be connected together, to make whatever it was you had in mind. This type is very much more simple and quicker to assemble.

These are the two fundamental types, and which one you decide to buy would depend on your present knowledge, and/or whether you wish to learn as you build. Some kits, especially some of the "little bits" types are very instructive, while the others enable you to put the equipment together in the shortest possible time. and with the least chance of errors in assembly.



The very comprehensive tool kit sold by Heath (Gloucester) Ltd. This firm market "Heathkit" kits.

TOOLS

It will be necessary for you to have some small hand tools, which should include, a small electric soldering iron, of about 25 watts with a 3/16 inch diameter bit, a pair of small wire cutters, a pair of small tapered nosed pliers, and one or two screwdrivers. Most of which with the possible exception of the soldering iron, you probably own already.

A multimeter is not absolutely essential, but it is highly desirable. This piece of equipment is moderately expensive (about £8 for a reasonable meter), but with care, should last for years. It would justify itself during its lifetime, merely on the strength of the many checking jobs it

would simplify.

SOLDERING

Do not be afraid of that soldering iron. Soldering is not the exact science that it is sometimes made out to be, and with modern resin cored solder and the correct iron, for the type of work you would have to do, nothing could be simpler, providing you use a little care and common

When making soldered joints, there are three

golden rules. First, make sure that the two pieces of material you wish to solder together are clean. That is, free from grease and free from insulation material like wax, enamel or any form of plastic. The second rule is never ever use any solder or flux on any electronic equipment except the resin cored type specifically sold for the purpose.

Thirdly, put the solder on to the joint, and not on to the iron. Allow the solder to flow around the joint, but be very careful not to overheat the components you are soldering.

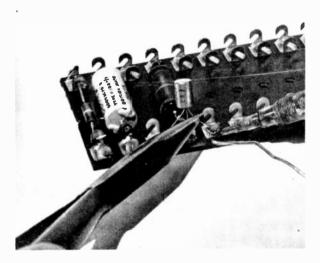
It is better to use a hot iron "quickly," than a cooler iron which must be held onto the joint for a longer time, thus allowing the heat to travel along the wire to the component, which may become almost as hot as the solder and be permanently damaged. This is very important with transistors.

It would be a good idea, if necessary, to practice making soldered joints with odd lengths of wire, before you start assembly of your kit.

HEAT SHUNT

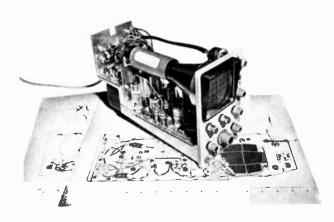
Using a "heat shunt," your long nosed pliers, held on the component lead between the component and the soldering iron, to keep the heat away from the comparatively fragile component, will save a lot of worry. A heat shunt should always be used when soldering transistors.

When completed, the soldered joint should be bright and shiny, and not crystalline as it would be if you moved the two parts before the solder has had time to cool and set. The photograph shows how to use a heat shunt and a good finished joint.



Photograph showing how to use a pair of pliers as a heat sink for soldering transistors.

Everyday Electronics, August 1972



A rather more ambitious kit, the Heathkit service oscilloscope. A large saving can be made by constructing your own test equipment of this type.

HIGH VOLTAGE

It is as well at this stage, to realise that there is some element of danger in most electrical or electronic equipment; if you are worried about building mains powered equipment, then all is not lost. There are still the transistor radio kits for you to build.

These transistor radios often provide added facilities not found on commercial types and are in every sense portable, as well as being free of high voltages. One of the many small radio kits available would make a good first project to get you used to the techniques.

COST

There would seem at first sight to be a very large difference in the cost of kits from the various manufacturers. The reason for this apparent difference can often be seen in the manuals provided with the kits.

Some manufacturers supply a very comprehensive book, which will provide all the information you could possibly need, both in the theory and the practice of building their kit. This applies particularly to the "little bits" kits, although there are some kits which are obviously intended for those who have already a working knowledge of electronics, and thus have only a limited amount of information.

The module type kit will generally provide a book or data sheet that will tell you what is required to connect up the separate panels, and a few do give a basic description of how the equipment works, but not a full description of the theory.

So, before you buy, ask to see the manual, and see if you can understand what is expected of you. You do not have to understand how the equipment functions, but you must understand how to put the parts together. Generally, the more expensive kits have the more informative manuals.

AFTER SALES SERVICE

Some suppliers provide a repair service for their customers, but if you can understand what is needed you should not need their help, with a consequent saving in time, effort and money. You will also get greater enjoyment from having built the equipment unaided. Nevertheless, check that your kit could be repaired by the supplier, should you have trouble.

It would be advisable, if you have never built any electronic equipment before, to start with a small kit, as previously indicated. The confidence you will obtain, plus the experience, will stand you in good stead if and when you want to build something more ambitious. But whatever you do, do not rush the job.

UNENDING

There is one snag; the process seems to be never ending, when you have improved your record player, you will feel that you ought to bring your radio up to the same standard, and then your loudspeaker system up to a quality that will do justice to the equipment you have already built. Or you might wish to construct some test equipment for your workshop, the possibilities are endless.

You can spread the cost, and for that matter the time, over as long a period as you wish.

Finally, comes the compensation for all your effort, when you have built it, whatever it may be, and your friends show their interest, you will have the satisfaction of telling them not where you bought it, but who built it.



The Radio and TV Components Unisound stereo record player. This kit is of the module type which can be assembled in a relatively short time.



Since we seem to have more than enough news of suppliers and new products this month we will quickly get down to buying problems for the constructional projects and then look at the news. There is just one point that has come to light from last months articles; the $0.5~\mu F$ tantalum capacitor specified for the Shaver Inverter is not readily available. A $0.47~\mu F$ type can be used and it does not in fact have to be tantalum.

Through The Lens Light Meter

Although the Through The Lens Light Meter is very simple to build, when properly calibrated and used it could prove invaluable. One important factor in the construction of this unit is the camera design and how it affects the layout of components in the case.

The case used for the prototype was originally a clear plastic box with hinged lid but any small box will do and some of the slide boxes supplied by film developing firms may be suitable.

Buying problems should be limited to the microswitch and the meter. Although there are many microswitches available a small one must be found if the case size is to be kept small. Different shops sell different types and if you want a small one unfortunately you will just have to look around until you find one—the current and voltage rating are not important in this design.

The meter is of course the most

expensive part of this unit and will probably cost about £2. One of the SEW MR38P, 100μ A types will be suitable. However, if you can find a small 100μ A meter (moving coil type) cheaper, then this would be suitable. The scale will have to be recalibrated so do not worry about how it is designated when you buy it.

Drill Speed Controller

No problems with buying for the *Drill Speed Controller*, the case can be obtained from most electrical suppliers as can the socket and blank panel used, if the supplier does not have these items he should be able to order them for you. It is worth noting that all the MK parts can either be obtained in white or cream and a better appearance will result if you get all one colour. Fixing screws should be supplied with the socket and cover plate.

The CSR (thyristor) used can be any 400V, 3A type if the CRS3/40 is not available. Just in case anyone is wondering, this unit is **not** suitable for use as a light dimmer but we may publish such a design in the future.

Weather Station

Not really any actual problems with buying for the Weather Station but a few points that must be noted. Firstly the two pots used for the wind direction vanes, these must be easy to turn, be of such a design that the stops can be removed and be of carbon construction—not wire wound. We have found that the plastic spindle types often fill all these requirements and two of these should be modified as indication in the article.

The motor used for the windspeed section can be any small model makers motor; the method of securing this to the case will vary according to the motor used. The three cup rotor can be fabricated from any suitably shaped items but a kit of all the rotor parts and the two windvanes is available from Kaspex, 16. Seymour Road, Tilbury, Essex RM18 7AP at a cost of 50p, including postage—mail order only.

When buying the moving coil meter look for the cheapest ImA design with a large scale; this scale will have to be redesignated for each range of the instrument. The case used for housing the monitor unit in the prototype was

purchased from G. W. Smiths but we believe a number of other firms can also supply this case.

New Products

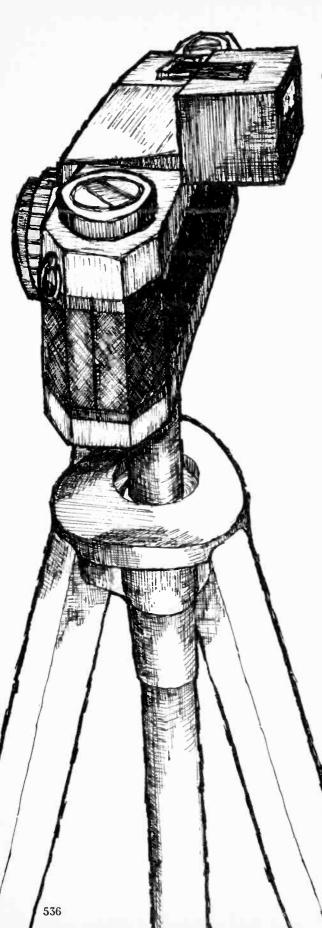
For all newcomers to stereo hi-fi reproduction. Linear have announced a low cost amplifier providing 5 watts output (they do not state if this is an r.m.s. or peak figure—or the speaker impedance used), with a frequency range of 20Hz, to 20kHz (this is not the same thing as frequency response). Price is £17.50 which seems reasonable but we would like to see proper figures quoted, not nondescript ones. Linear give a 12 month inclusive guarantee-more literature from your local dealer or Linear Products Ltd., Electron Works, Armley, Leeds, LS12 3SA.

A 12V Invader soldering iron has recently been introduced by Adcola products, who advertise in our pages. Designed to operate from a 12V car or boat battery, the iron is available in two sizes, 216 and 14 inch bit diameter, rated at 23 and 27 watts respectively. The iron is provided with crocodile clip connections, 12ft of lead and a fire resistant transit cover which fits over the element and bit, allowing the iron to be stored away without having to wait for it to cool down. The 316 inch bit model costs £2.37 and the 14 inch model £2.47.

Suppliers

Two news items concerning suppliers, the first, which is of general interest, is that G. W. Smiths and the Laskys group have merged. These are probably the two largest companies retailing hi-fi equipment and electronic components. It has been stressed by the officials of Audiotronic Holdings Ltd., a new company formed to amalgamate the two companies, that there will be no change in trading policy or shop names etc. A possibility arising from the merger is the opening of a new shop somewhere outside the London area, but this may be well into the future.

Of more direct interest to readers is the opening of a new shop by Henry's Radio Ltd. at 404-406, Edgware Road, London, W.2. This new branch should have been open a few weeks when you read this; it will carry mainly components plus audio equipment and test gear.



through

A simple light meter for use with single lens reflex cameras.

By E. B. Eves



Approximate cost of components £2.75 plus case

WITH the dramatic fall in price which has taken place over the last few years many more people own single lens reflex cameras than would have seemed likely even five years ago. As most of these cameras are fitted with interchangeable lens systems there is a much greater scope for using a camera for "unusual" photography, employing the use of extension tubes or telephoto lenses.

Many people use their cameras as an extension to another hobby. In this connection bird watching and other naturalist pastimes come to mind in conjunction with a camera and telephoto lens, while many collectors of small items such as stamps and coins, make use of extension tubes to take close-up pictures of their specimen pieces.

The main problem which arises in both these cases is that of finding the correct exposure. The technique of taking several exposures is suitable if cheap film is being used but becomes rather expensive if colour or special films are employed. The meter described below overcomes this problem, although it is only really suitable if the camera is being used on a tripod, as the viewfinder has to be obscured while the meter is in use.

Everyday Electronics, August 1972

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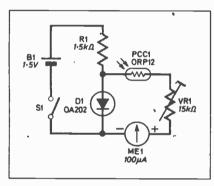


Fig. 1. Complete circuit diagram of the Through The Lens Light Meter.

THE CIRCUIT

The circuit is a simple one using a photo conductive cell, a meter, a silicon diode, a battery and two resistors, one fixed and one variable. The circuit is shown in Fig. 1.

To describe the action of this simple circuit let us assume that it is in two parts, the first consisting of Bl, Sl, Rl and Dl. It will be seen that with Sl closed, current will flow around the circuit since Dl is forward biased. In this condition the diode, which is a silicon type, will always have a voltage of approximately 0.6 volts across it.

It is this small voltage that is measured by the second part of the circuit which is a low range voltmeter formed by PCC1, VR1 and ME1. Once VR1 is set the voltmeter will read a value dependent on the resistance of PCC1; the voltage being measured is always steady at 0.6 volts. Since the resistance of PCC1 varies in proportion to the level of light falling on its surface ME1 will, in fact, indicate the light level.

THE CASE

The construction of the case depends on the type of camera to which the device is to be fitted. The case shown in Fig. 2 was designed to fit a camera where the accessory shoe is mounted directly above the eveniece.

The only requirement is that when clipped into the shoe the grommet surrounding PCC1 must be centred against the view-finder. The micro-switch S1 is depressed by the pressure against the camera and "automatically" switches the circuit on. The case used can be any plastic or metal case of minimum dimensions 2 inches by 134 inches by 134 inches but this size will depend on the meter available.

CONSTRUCTION

The components can be mounted as shown in Fig. 2. The photo-conductive cell must be provided with sufficient lead length to allow it to be positioned as described above.

Start construction by mounting S1, PCC1, ME1 and B1 inside the case, B1 can be mounted by a small Terry clip with a flat brass connector at each end, held in place with foam rubber glued to the side of the case.

Next attatch Rl and VRl as shown, connect up Sl and PCCl and finally solder in diode Dl using a pair of long nose pliers held on the wires to prevent heat from the soldering iron damaging the diode.

COMPONENTS

The photo-conductive cell PCC1 is an ORP 12 or similar type. The meter can be any small inexpensive 100μ A moving coil type. The front of the meter is removed to allow a paper scale

to be fitted over the existing one.

Alternatively if some care is taken a slot can be cut in the upper edge of the front cover, when it is removed, using a hot sharp knife blade and finishing with fine sand paper, in such a way that a very thin piece of aluminium, sprayed with white paint and calibrated (on both sides if required), can be slid in over the top of the existing scale and beneath the pointer of the meter, Fig. 3. By this means scales calibrated for different film speeds may be prepared and fitted easily. The method of calibration will be described later in the text.

The variable resistor should be a 15 kilohm carbon skeleton preset type mounted in such a way that it can be set by means of a screw-driver

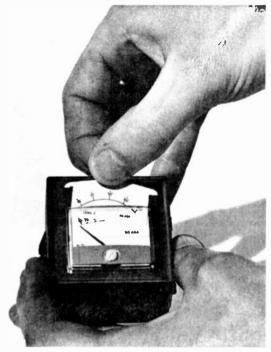


Fig. 3. Method of inserting various scales in the meter movement.

SETTING UP AND CALIBRATING

When the circuit has been wired up and checked the photocell should be pointed into a fairly dark corner and the micro-switch pressed. The meter should give a reading and moving the unit about should change this reading as the light level changes. If this is not the case the circuit should be re-checked. The light meter may now be mounted on the camera which should be set on a tripod with the normal lens fitted, pointing at a plain white surface.

The camera need not be focused, indeed, if the surface is not uniform the camera being out of focus as far as possible would be an advantage. The white surface should be illuminated, preferably by normal daylight. The meter will give a reading dependent on the light level.

An ordinary light meter is used to determine the light level from the surface at the camera. The camera is then set up using this reading. at ¹100 of a second, for the film normally used.

The needle on the light meter is set to centre scale by means of the resistor VR1. The point at which the needle rests must be marked 1₁₀₀. If the aperture is now opened by exactly one stop the needle will move, its new position should be marked 1₂₀₀, if closed to one stop below the original the scale must be marked 1₅₀. This is repeated at one stop intervals until a scale as shown in the photograph is obtained.

The scale is now calibrated for the chosen film speed, other scales may be made by setting the camera to a chosen aperture and marking the scale with the corresponding shutter speed, the variable resistor must not be altered after the initial calibration, and can be sealed with a spot of candle wax dropped, hot, onto the centre of the preset.

If two film speeds are normally used the scales for them may be marked on one scale.

USING THE METER

When the meter has been completed an easy check can be carried out to see if the calibration is correct. The camera is mounted on a tripod with the normal lens fitted and a meter reading of the exposure required is taken by means of a normal light meter. The meter described above is then fixed to the camera and the aperture adjusted until each speed in turn is reached. The corresponding aperture should coincide with the required exposure as found by the other meter.

In general, to use the through the lens meter a shutter speed is selected, the camera focused on the subject and the aperture adjusted until the needle rests on the selected shutter speed

Components..



Resistor

R1 $1.5k\Omega$ $\frac{1}{4}W \pm 10\%$ carbon

Variable Resistor

VR1 $15k\Omega$ skeleton preset

Diode

D1 OA 202 or any small silicon diode

Miscellaneous

PCC1 ORP 12 photo-conductive cell

ME1 100µA moving coil meter, small type

B1 1.5V HP7 battery

S1 S.p.s.t. miniature micro-switch or

small push button

Case (see text), grommet to hold PCC1, wire, metal for battery connectors and mounting bracket, fixings for mounting bracket, small Terry clip

through the lens light meter

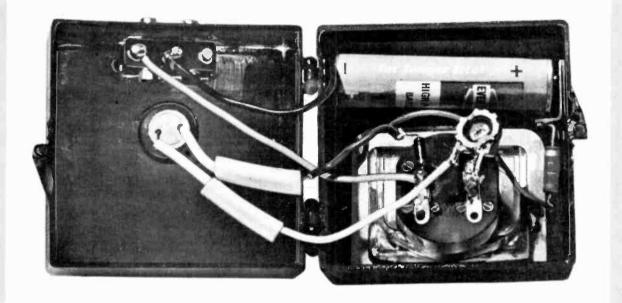
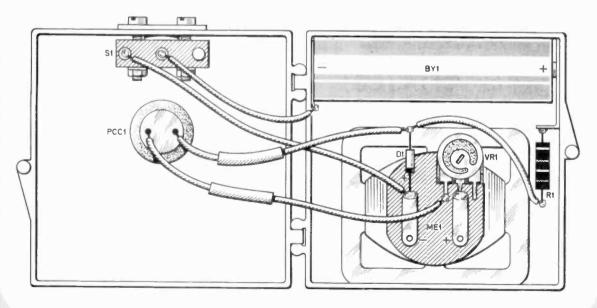
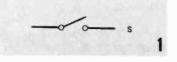


Fig. 2. Complete construction and wiring details of the Through The Lens Light Meter.

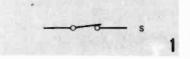


P guide to circuit

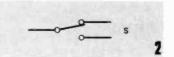
Switches



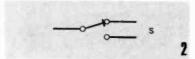
Single pole single throw make contact



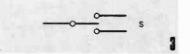
Single pole single throw break contact



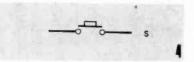
Single pole double throw changeover break before make



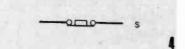
Single pole double throw changeover make before break



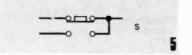
Single pole double throw centre off



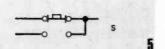
Single pole push to make contact



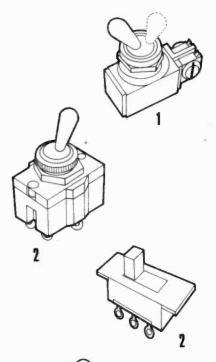
Single pole push to break contact

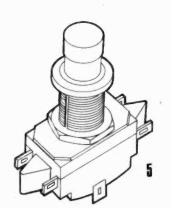


Single pole push button changeover break before make

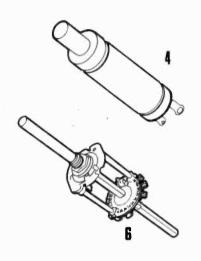


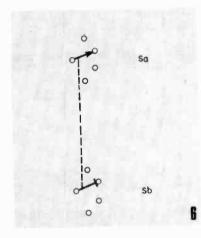
Single pole push button changeover make before break





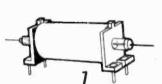
symbols...part 3





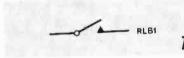
Double pole multi-way switch (usually rotary). Having one break before make pole and one make before break pole. Each pole has a suffix letter after the switch number. The dotted line indicates that two or more poles are operated simultaneously



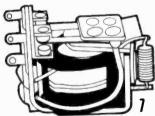




Relay coil with resistance indicated in ohms. Relay coils are annotated RLA, RLB, etc. The number of contact sets operated by the coil is shown under the coil annotation e.g. RLB



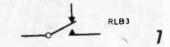
Make contact



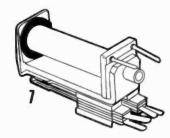


Break contact





Changeover break before make contact





Changeover make before break contact



Changeover, both sides stableno normal rest position.



TEACH-IN ... FOR BEGINNERS

By Mike Hughes M.A.

11) REACTANCE & INDUCTANCE

This month we'll begin with an experiment to illustrate a.c. resistance. You will need a reasonably high resistance a.c. voltmeter. This is to show that although a capacitor will block the flow of direct current, alternating current appears to flow.

Make the circuit of Fig. 1 and measure the r.m.s. a.c. voltages at points A and B. You should find that (making allowances for experimental error and component tolerances) that the a.c. potential at point B is approximately half that at point A. This means that current must be flowing "through" C1 otherwise the potential at B would be the same as at A.

As the potential we measure is approximately half the supply voltage, this means that the resistance of Cl to the flow of current must be approximately half the total resistance of Rl and Cl in series.

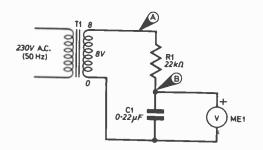


Fig. 1. Circuit diagram for experiment to illustrate capacitor reactance and impedance. Capacitor C1 must not be an electrolytic. Try different values of C1.

AC RESISTANCE

Resistance to a.c. by a capacitor is not a simple thing like that of a normal fixed resistor, because it will vary with frequency. Its effect is very similar and hence we measure it in ohms, but, we also have to state at what frequency we are quoting the value.

To distinguish it from conventional resistance we call this effect of the capacitor, "reactance" and it can be fairly easily calculated for a given value of capacitance at a certain frequency from the following expression:

Reactance of capacitor
$$X_{c_1} = \frac{1}{2\pi \times f \times C}$$

Where X_c is in ohms, π is the constant 3·142, f is the frequency in Hz and C is the value of the capacitor in farads.

Thus we can calculate the reactance of our $0.22\mu F$ capacitor at 50Hz:

$$X_{c} = \frac{1}{2\pi \times 50 \times 0.00000022}$$
= approx 14,000 ohms or 14 kilohm

We therefore say that the reactance of a $0.22\mu F$ capacitor at 50Hz is approximately 14 kilohms.

Do you notice, though, that this is much less than half the total resistance of R1 and X_c added together. When dealing with capacitors and resistors in series we cannot simply add resistance to reactance to obtain what you might call the total resistance. We have to do it in a much more roundabout fashion and the end result

(although still measured in ohms) is called the "impedance" and is usually given the symbol Z.

For a single resistor and single capacitor in series the impedance Z is calculated as

Impedance,
$$Z = \sqrt{(R^2 + X_c^2)}$$

where Z, R and X_c are measured in ohms.

This means that the impedance Z of our circuit is:

- $= \sqrt{(22,000 \times 22,000 + 14,000 \times 14,000)}$
- $=\sqrt{(484,000,000+196,000,000)}$
- $= \sqrt{680,000,000}$
- = approximately 26,000 ohms or 26 kilohm

You can now see that the reactance of our capacitor is just over half the total impedance of the circuit and this is why the r.m.s. voltage we measured at point B was approximately half the supply voltage.

FERRITE TRANSFORMER

Last month we saw that when we connected a d.c. source of current across the primary side of a transformer, we got a momentary voltage generated across the secondary. We must now look at this a little closer and see which way the current in the secondary flows when the primary is connected a given way round.

It is a little difficult to do this with the Friedland transformer because it is not obvious which way the turns go round the core. To simplify matters it is best if we make a very crude transformer for ourselves.

For this you will need about 6 inches of $^{3}_{8}$ inch diameter ferrite aerial rod and some 28 or 30 swg enamelled copper wire (a 2oz reel of this will be ample).

CONSTRUCTION

Anchor the free end of the copper wire to the ferrite rod with Sellotape—leaving about 6 inches flying for later connections—and wind 200 turns of wire over a length of about 1 inch. The turns can be wound on top of each other. When you have done this anchor the last turn in place and cut the wire—again leaving about 6 inches free.

Leave a gap of about 12 inch and wind an identical coil on the same rod adjacent to the first but make sure that the turns go round the rod exactly the same way as before, see Fig 2(a). We now have a simple transformer with a turns ratio of 1:1. We must hasten to add that this assembly is only applicable to our experiment and you should not use this technique for making a power or mains transformer!

Fig. 2(b) is the schematic of this transformer; notice the black spots at one end of each of the windings. These denote the wires leading into the "start" of each coil (i.e. the wires you anchored to the ferrite rod as you started to wind each coil). It is most important in this

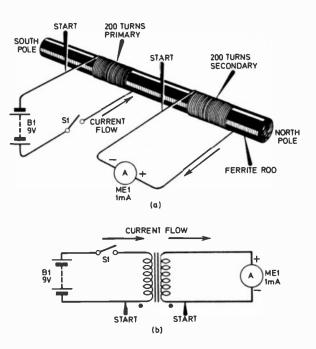


Fig. 2(a). The experimental transformer wound on six inches of ferrite aerial rod. (b) Shows the schematic diagram with designations for the start of each winding. Arrows show direction of current flow when the circuit is "made". The current through the secondary reverses when the primary circuit is broken.

experiment that you can recognise these leads.

Before proceeding further you must remove the insulating enamel from the four ends. This can be done by scraping gently with a razor blade or rubbing with fine emery cloth.

Now decide which of the two coils you will call the primary and which the secondary, and then connect the secondary across the 1mA meter on the Demo Deck with the **start** end going to the **negative** terminal. The negative terminal of a 9V battery should be connected to the start of the primary.

When you complete the primary circuit, by connecting up the positive terminal of the battery, watch the meter very carefully. You will see a very small kick in a positive direction which very rapidly falls back to zero. Now disconnect the battery and you get a slight "kick" in the negative direction—just as we had last month only the "kicks" are not so strong.

INDUCTION-MUTUAL INDUCTANCE

Now we know the sense of the windings we can see that a change in current flowing one way round the core in the primary gives rise to a build up of magnetisation of the core (for the direction of current flow we are considering the north pole of the magnet is as shown in

Fig. 2) which in turn causes a momentary current to flow in the secondary—but in the opposite direction around the core. This effect is called "induction".

We say that the "signal" in the secondary is "induced" by the change of current in the primary. If the secondary coil was connected to a very fast reading voltmeter and we were able to control the rate of build up of the primary current, you would see that the output voltage was proportional to the rate of change of primary current.

Secondary voltage, $V = M \times \text{Rate}$ at which primary current changes

V is measured in volts, the primary rate of change in amperes per second, and M is a constant of proportionality set by the numbers of turns on each coil, the distance between the coils, the dimensions of the coils and the magnetic properties of the core.

We call the constant, M, the "mutual inductance" (between the two coils in question). This is measured in units called henries (abbreviation "H") and frequently we come across mH (millihenries) and μ H (microhenries).

One henry is the mutual inductance which will cause an output voltage of one volt when the primary current changes at the rate of one ampere per second.

SELF INDUCTANCE

Now imagine a single turn on the primary and a single turn (having the same winding sense) on the secondary—Fig. 3.

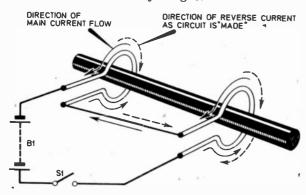


Fig. 3. The main current from the battery through the first turn induces a reverse current in the second turn as the circuit is "made". This resists the build-up of the main current, this effect is called reactance. The influence of the turns on each other is called self inductance.

Let us connect the first turn to the second turn (exactly as if we had wound on two adjacent turns) and pass a current through both. The change in current through the first will still induce a reverse current in the second and this induced current will oppose the flow of the original current—because the two turns are connected in series.

This means that as we try to generate a fast change of current the inductance between the two turns will try to oppose it; this slows up the rate at which our "primary" current can build up. In effect the two turns work together to resist the flow of current—especially if we try to make the current change fast. The faster the change in current we apply the stronger will be the opposing current and hence this resistance.

This obviously is not a simple sort of resistance because it is generated by the "self induction" between the two turns and has nothing directly to do with the material from which the wire is made.

As with capacitors rates of change affect the degree of this resistance and hence we call the effect reactance. It is still measured in ohms with a symbol $X_{\rm L}$. To differentiate between mutual, and self inductance we designate the term L to the latter. It is still measured in units, or fractions, of henries and is dependent on the number of turns, their spacing, diameter, and the properties of the core. The higher the self inductance the greater the reactance of the coil for a given rate of change of current.

An inductance may have a very low resistance to direct current, but if we try to pass alternating current its resistance increases—we call this form of resistance "reactance" and this increases with frequency.

We can calculate the reactance of a coil if we know its inductance from the following expression

$$X_1 = 2\pi f L$$

where X_L is the reactance measured in ohms, π is our old friend 3·142, f is the frequency at which we want to know the reactance, and L is the inductance in henries.

If we have an inductor in series with a resistor in a circuit we cannot simply add the value of the resistor to the reactance of the inductor to obtain the total resistance to a.c. current flow. As with the capacitor, we have to calculate the impedance. The formula is very similar:

$$Z$$
 (impedance) = $\sqrt{(R^2 + X_L^2)}$

SERIES AND PARALLEL

If you have several inductors in series (provided they are physically far enough apart to prevent mutual inductance) you can simply add the values of inductance together to find the total effect:

$$L_{\text{total}} = L_1 + L_2 + L_3 + \ldots +$$

For inductors in parallel

$$\frac{1}{L_{\text{total}}} = \frac{1}{L_1} + \frac{1}{L_2} + \frac{1}{L_3} + \dots +$$

HIGH VOLTAGE

If we have a large inductor in transistor circuits that switch currents on and off very fast we can sometimes have a problem.

If you pass a reasonably heavy current through an inductor and then break the circuit very quickly, the energy that is stored within the magnetised core will fall and while doing so will try and generate a reverse current; but as the circuit is broken the current cannot flow anywhere and we can momentarily generate a very high voltage across the ends of the inductor.

We can do a simple experiment to demonstrate this. You will need the Friedland bell transformer a 9V battery and a 60 to 70V neon bulb (it should be a "bare" neon and not the type with a built in resistor as used for panel indicators).

Connect the neon across the primary (mains) side of the bell transformer and then connect the battery up across the same terminals—do not connect anything to the secondary, see Fig. 4.



Fig. 4. Experiment to show that a momentary high voltage can be developed across the terminals of a high value inductor when the current suddenly stops flowing.

The neon does not light up because we are only supplying 9V to the circuit and the bulb needs at least 60/70V across it before it will glow. Now quickly disconnect one terminal of the battery and watch the bulb, but do not hold the bare wires, otherwise you will get a small electric shock; you should see a brief "flash" indicating that for a fraction of a second the voltage across the ends of the transformer primary (our indicator), must have risen to a value greater than 70V.

If you could make and break the circuit very quickly you would see a steady glow from the neon. You can do this by using a buzzer that runs from batteries. Connect the neon directly across the coil and connect the battery.

You should see a steady glow from the bulb indicating the continuing recurrance of the high voltage caused by the inductance of the buzzer's coil and the constant making and breaking of the circuit by the contacts. Incidentally it is this high voltage that causes the sparking across the gap of the contacts.

Sometimes you will see the specification for a switch that says it may only be used to switch d.c. with non-inductive loads. This means that the contacts are not made to withstand the sparking that can be caused.

DIODE SHUNT ACROSS RELAY

Quite often a transistor is used to control the current flowing through the coil of a relay—a typical circuit is shown in Fig. 5.

If the transistor is made to switch off very fast by removing the base current, the potential across the coil will momentarily shoot up to a value much greater than the supply voltage in such a way that the potential at the end connected to the collector of the transistor will go more positive than the positive rail. In severe cases this might momentarily exceed the collector base reverse breakdown voltage and the transistor might be destroyed.

To prevent the collector going any more positive than the positive rail, designers sometimes incorporate a diode across the relay coil. This diode has no effect on the normal working of the circuit but will act as a short circuit to reverse voltages generated across the coil.

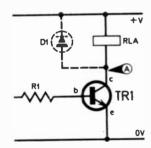


Fig. 5. When a transistor switches off very quickly, the voltage at point A can rise to a positive value much higher than the supply voltage. Sometimes a diode, D1, is connected across the relay coil to "short circuit" this high voltage pulse and thus protect the transistor.

PHASE SHIFT

There is another very important effect which we have not yet covered. This is to do with the fact that we cannot simply add reactance to resistance to find the total effective resistance of a circuit.

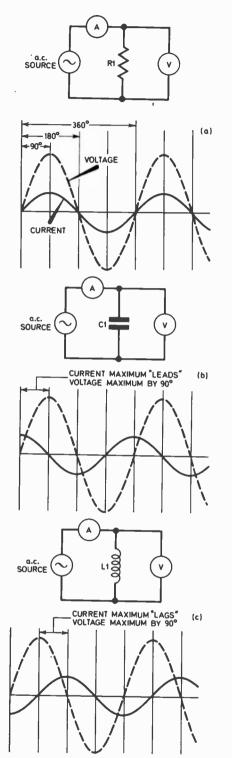


Fig. 6. Voltage and current phase relationships. (a) For a resistor, current and voltage are in phase. (b) For a capacitor, the current flowing "through" it leads the voltage across it by 90 degrees of angle $(\pi/2)$. (c) For an inductor, the current flowing through it lags the voltage across it by 90 degrees $(\pi/2)$.

In Fig. 6(a), (b), (c) we show a resistor, a capacitor and an inductor respectively connected across a source of a.c. voltage. In the case of the resistor we can say that at any instant in time the current flowing through the resistor is directly proportional to the voltage; i.e. when the voltage is maximum positive the current is maximum positive, when the voltage is zero the current is zero.

This is shown graphically by the superimposed voltage and current waveforms. We say the voltage and current are "in phase".

When the voltage across the capacitor is zero, but rising from negative to positive, most current will be flowing into it, but as the voltage across it reaches maximum (i.e. the capacitor is fully charged in one direction) the charging current falls to zero.

As the voltage falls from maximum positive towards zero, the capacitor in effect discharges and current starts to flow in the opposite direction (i.e. becomes negative).

We can say that as the voltage passes through zero but is rising in a positive direction we get maximum positive current; when the voltage reaches maximum we get zero current and when the voltage passes through zero in a negative direction we get maximum negative current. This means there is a shift in phase between the maximum current and voltage.

The shift is a quarter of a wavelength which is 90 degrees or 7 radians. We say by definition that the current "leads" the voltage by 90 degrees.

The male and female approach works here because the opposite happens with an inductor. We get maximum rate of change of current when the current waveform passes through zero. If the current is rising in the positive direction this gives us maximum postive voltage across the inductor. As the current passes through maximum, the rate of change momentarily becomes zero (it stops at the maximum before coming down again) and this gives rise to zero voltage. As soon as the current passes through zero in a negative direction we get maximum negative voltage, and so on. Again there is phase shift between the voltage and current but this time the current "lags" behind the voltage by 90 degrees.

RESISTOR—CAPACITOR—INDUCTOR COMBINATION

The reason why we could not simply add resistance to reactance for a resistor/capacitor or a resistor/inductor circuit was because of the relative phase shifts between voltage and current waveforms. Likewise there is a phase shift between an inductive and a capacitive circuit only more so—180 degrees—so therefore we cannot simply add together the reactance of a capacitor to that of an inductor in series.

Capacitors

A capacitor will not pass d.c.

A capacitor passes high frequency a.c. easily The reactance of a capacitor decreases as the frequency increases

Inductors

An inductor passes d.c. easily An inductor is a very poor conductor of a.c. The reactance of an inductor increases as frequency increases

Formula for reactance

$$X_L = 2\pi f L$$

Series connection

$$L_{\text{total}} = L_1 + L_2 + \dots$$

Parallel connection

$$\frac{1}{L_{\text{total}}} = \frac{1}{L_1} + \frac{1}{L_2} + \dots$$
High inductance gives high reactance

 $C_{\text{total}} = C_1 + C_2 + \dots$

 $\frac{1}{C_{\text{total}}} = \frac{1}{C_1} + \frac{1}{C_2} + \dots$

 $X_{\rm c} = \frac{1}{2\pi fC}$

High capacitance gives low reactance

Strangely enough (and this is difficult to prove without fairly complicated mathematics) we have to subtract the reactance of one from the other.

The formula for calculating total circuit impedance becomes:

$$Z = \sqrt{[R^2 + (X_L - X_c)^2]}$$

Next month: Amplification. Additional components required for next month: microphone insert type ACOS MIC 43-3 or Duvidal CM20; resistors: 3.3 megohm (1 off), I megohm (I off), 330 ohm (I off).



Ruminations By Sensor

Summer Is Icumen In

I have just finished a spell of home decorating-three weeks of painting and papering-which has stopped me from completing any of the four electronic construction projects that I have on hand at present. For me, home decorating is a continuous process with long troughs of low level activity and occasional high peaks of short duration. It is one of the latter that has just been completed.

My gardening follows a similar pattern, so that the "seed-time and harvest" of the good husbandman could be more accurately described in my garden as "weedtime and putrefaction". Since I bought an electric lawn mower my schoolboy son mows the lawn regularly and I recommend other fathers to follow my action.

Weeding is a problem to which I have not found an answer, apart from getting down onto one's knees and digging them out or preferably, getting one's wife to do it (and there's nothing novel about that!). I use various weedkillers and find them effective but they have their limitations and bring their own problems.

Back To Work

So; having, as it were, subcontracted most of the gardening, I folded up the pasting table and turned again to the bench. What a mess! Why does my bench gather all the family's rubbish?

The surface was covered with items ranging from eggs (a pullet's first and a goose's), an old blackbird's nest, two strings of beads, seashells, an old teapot, through a variety of household utensils needing attention to a couple of electric motors awaiting assembly. The whole was overlaid with a fine layer of sawdust, ac-

quired during my last session with the circular saw. My own neglected projects were not in evidence-a second layer had built up over them.

Have you noticed that if you leave a job undone long enough the need for it often disappears? Indeed, it can become impossible to carry on with it due to the parts becoming obsolete and unobtainable-and that doesn't take long these days.

Well, then, what is to be done about those unfinished jobs? I think that one must be ruthless; remove from the bench all things not immediately required (put them into the children's bedrooms), select one job that is possible to finish and concentrate on that, do not allow yourself to be persuaded into anything else, be single-minded to the point of selfishness. Set yourself a reasonable time for completion and stick to it.

If you find the method works, let me know, perhaps I'll try it myself!

WEATHER STATIONS

by D. T. Goodwin B.Sc.

A simple device using basic electronic / principles for measuring some basic / weather parameters.

ow convenient it would be to know the weather details at the push of a button (or the throw of a/switch), whilst sitting comfortably at home. The complete Weather Station is in two parts, that is to say, it is built in two separate cases, one (the transducer case) to be installed outside in an unskeltered unobstructed position, and joined by several wires or multicore cable to the monitoring case located inside the house or laboratory.

CIRCUIT'S

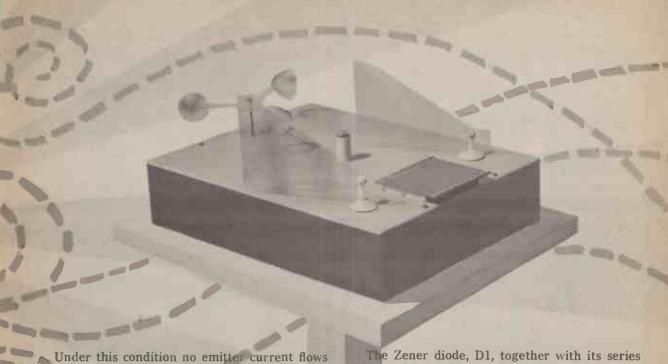
The complete circuit diagram for the Weather Station is given in Fig. 1. It is seen to consist of five separate and independent units wired together around a three-pole-four-way wafer switch S2. These units are: rain indicator, temperature monitor, wind speed and direction indicator, and a light intensity indicator.

In the prototype, with the exception of the rain indicator, the units share the meter ME1 via switch S2 which is used to select the parameter to be measured.

RAIN INDICATOR

The rain indicator circuit has been designed so that when rain or moisture falls on the Veroboard sensor, the lamp LP1 on the control box front panel lights up. The components of the rain indicator consist of TR1 and TR2 directly coupled, R1, R2, LP1 and the Veroboard sensor.

With S1 switched on, and under dry conditions (no moisture on the sensor) the base of TR1 is virtually open circuit i.e. no base current flows.



Under this condition no emitter current flows in TR1, and hence no base current in TR2. Transistor TR2 is thus off—no current flows through the lamp.

When moisture (rain) bridges the sensor, the very him resistance between the adjacent strips of the sensor is greatly reduced and base current flows in TR1, allowing emitter and collector current flow.

This condition allows base current to flow in TR2 thus turning the latter on and illuminating

the lamp.

If it is found that the sensitivity of the sensor is too high, it can be reduced by inserting a 5 kilohm preset potentiometer (wired up as a variable resistor) in series with R1; adjust to the required sensitivity.

With no moisture on the sensor, the unit is very economical, current drain from the battery being minimal and due only to the leakage of

the transistors.

However, when it is raining, current drain is considerable and for this reason it is advisable to switch S1 off.

If desired, the lamp may be replaced by a suitable relay that can be used to switch on an audible alarm.

TEMPERATURE INDICATOR

Ambient temperature is indicated on ME1

with S2 in position 1.

The unit utilises the variation of resistance with temperature in a device called a thermistor. The type used in the prototype is one whose resistance decreases with increasing temperature. It is used, with R4 to form a potential divider chain.

The Zener diode, D1, together with its series resistor R3, produces 3·3 volts across the transistor and the potential divider chain. This lower voltage helps to reduce the effect of self-heating of the thermistor.

As the temperature increases, the resistance of RTH1 drops and more base current flows in TR3, and therefore, more emitter and collector current flows. The collector current flows through ME1 and can be varied by VR1.

To calibrate the temperature scale, a conventional thermometer is needed (unless it is only coarsely calibrated as in the prototype as "cold" "warm" "hot" and "very hot," by trial and error

on days of these descriptions).

If we are only concerned in measuring outdoor temperatures, then 40 degrees Centigrade is a suitable figure to make as our full-scale deflec-

tion on ME1 (in the U.K. that is).

Place the conventional thermometer alongside RTH1 on the "transducer box" with S2 in position 2, and bring up a heat source in the vicinity of the thermometer and RTH1 (the bulb of a table lamp is a good source of heat) and adjust VR1 so that the meter reads full scale when the thermometer indicates 40 degrees Centigrade. Mark this position on the meter scale with 40. Repeat this procedure working down in five degree steps.

WIND SPEED

The wind speed is indicated on ME1 with S2 in position 2.

This is quite a straightforward device that uses a small d.c. motor rated at between 3-6 volts as a generator—this was found to be satisfactory in the prototype.

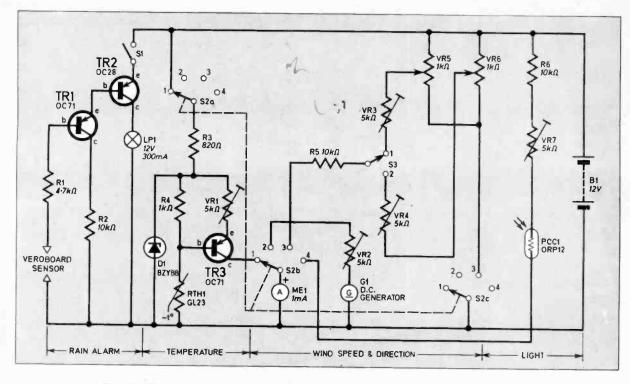


Fig. 1. The complete circuit diagram of the simple Weather Station.

The rotor containing the wind cups is connected to the shaft of the generator, and as the wind carries the cups round, a current is generated which passes through the meter via VR2.

The variable resistor VR2 controls the current flowing into the meter and should be adjusted such that when the rotor is spinning at maximum wind velocity the current flowing is 1mA.

The meter scale can be calibrated in several ways: in the prototype it was carried out by experiment on days of different wind strength, i.e. "light," "blowy," "strong," "gale," and the face marked accordingly.

Alternatively, if an accurate method of producing or calculating a wind speed in m.p.h. is known (such as a wind tunnel or mounting on a car and driving at an observed speed on a still day) the scale may be calibrated in m.p.h.

WIND DIRECTION

Wind direction is indicated on ME1 with S2 in position 3.

The combination of R5 with each of VR3 and VR4 in series with the meter converts ME1 to a voltmeter than can be adjusted (by means of VR3 and VR4) to read 10-15V full scale.

The full battery supply of 12V is across VR5 and VR6. Depending on the position of S3, the meter records the voltage at the wiper of VR5 or VR6 relative to the negative line.

The wind direction vanes are connected to the spindles of VR5 and VR6.

Two vanes (with potentiometers) have been used to eliminate the small portion on the potentiometers which will not give a reading when the stops have been removed, see later.

The two potentiometers are mounted in approximate opposite directions to each other. Only one is used at a time, S3 being employed to change from one to the other when one goes into the null region. In fact, it is advisable, for accurate measurements, to switch S3 when the needle is near to either end of the direction indicator scales.

There are two scales, one for each switch position and these should be calibrated as follows; for the upper scale (1), set S3 in position 1, and VR3 to maximum resistance. Position VR5 vane so that ME1 reads a maximum. Adjust VR3 so that this maximum is full scale.

Repeat this for the other vane using VR4 with S3 in position 2. Return S3 to position 1.

Rotate the vane on VR5 so that a small fraction above zero is shown on ME1. Mark this point on the meter face with one of the four compass points, i.e. N, E, S or W—say N for example.

Now turn the vane through 90 degrees clockwise and mark the new needle position E. Rotate a further 90 degrees and mark S. Another 90 degrees rotation gives the W position.

Turn the vane back to the S position.

Switch S3 to position 2, and position VR6 so that the needle lines up with the N position on the upper scale with the vanes in the same

Components....

Resis	stors	1/
R1	4 · 7kΩ	V
R2	10k Ω	,
R3	8200	\checkmark
R4	1kΩ	
R5-	10kΩ	
R6	10kΩ	

All # watt ±10%

SHOP TALK

Potentiometers

VR1 5k(2 skeleton preset VR2 5k(2 skeleton preset VR3 5k(2 skeleton preset VR4 5k(2 skeleton preset VR5 1k(2 finear carbon VR6 1k(2 finear carbon VR7 5k(2 skeleton preset Light Dependant Resistor PCC1 ORP12

Semiconductors

TR1 OC71 germanium pnp
TR2 OC28 or AD142 or similar
TR3 OC71 germanium pnp
D1 BZY88 C3V3 400mV Zener diode

Thermistor

RTH1 GL23 glass bead type

Switches

S1 S.P.S.T. toggle or slide S2 Three-pole-four-way wafer S3 S.P.D.T. toggle or slide

Miscellaneous

ME1 1mA meter LP1 Lamp 12V 300mA plus panel type holder

G1 Generator 3-6V d.c.—see text B1 Battery 12V—use PP1 6V (2 off in

series)

Veroboard 0.15 inch matrix: control box 30 x 10 holes, rain sensor 24 x 20 holes (approx.); Rotor cups and vanes; 1/4 inch clear Perspex or similar material; pointed knob; materials for transducer box; control box case, 5 x 5 x 8½ inches with sloping front panel.

direction. Mark S on the lower scale (2), and then carry out similar rotations to those above for marking the lower scale W, N, and E respectively. Secure VR5 and VR6 in these positions.

LIGHT LEVEL INDICATOR

The light level is indicated on ME1 with S2 in position 4.

The circuit utilises a photoconductive cell, PCC1, whose resistance varies with light intensity from approximately 10 megohm in dark conditions to about 75 ohm in bright light.

As the light intensity varies so the total resistance of the circuit R6, VR7 and PCC1 in series

Everyday Electronics, August 1972

varies, affecting the magnitude of the current flowing through ME1.

Once again, the maximum current flow should be 1mA. To realise this condition VR7 should be set so that 1mA flows when very bright sunlight is incident on PCC1.

To do this set VR7 at a maximum (i.e. 5 kilohm) and point PCC1 at the sun on a clear summer day. Now adjust VR7 so that the reading is full scale.

The full scale position should be marked "very sunny" and the zero position "dark".

For intermediate intensities it is suggested that these be marked "dull," "bright," "sunny," the positions being determined by days when these light intensities are evident.

METER DISPLAY

A suggested meter-scale is shown in Fig. 2. It consists of five bands to represent each of temperature, light, wind speed and wind direction (upper and lower which are used in conjunction with S3, see above).

It is suggested that these concentric bands are all firstly drawn together on a piece of plain paper and then pasted in position over the original scale.

Pencil markings will be sufficient to locate, positions which can then be inked over neatly or printed with Letraset.

The start and end positions of the new scale must coincide with the original "zero and full scale" positions of the meter.

CONSTRUCTION-CONTROL BOX

The layout of the components on the Veroboard is shown in Fig. 3.

The circuitry in the control box is built on a/piece of 0.15 inch matrix Veroboard size 30 by

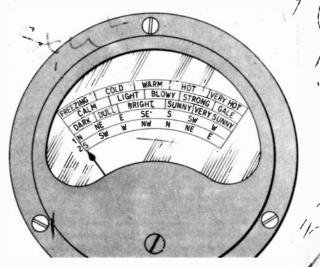


Fig. 2. A suggested meter scale display as used in the prototype. The individual scales will have to be calibrated as described.

551

10 holes with the copper strips cut as detailed.

Begin assembly by fixing TR2 in position and carefully soldering. The collector of TR2 is its casing and this connection is made via a small nut and bolt through a solder tag. This connection is also made through the Veroboard at position G1/F1 by drilling a hole to take the bolt used. Continue by soldering the resistors and potentiometers in places as shown in Fig. 3.

Remember that the potentiometers are being used as variable resistors therefore only one outside leg is used with the middle leg. The redundant leg should be bent upwards away from the board, making sure neighbouring components are not touched.

Now attach the switches, S1, S2 and S3, the lamp holder and bulb, and the meter to the front panel as shown in Fig. 4, and wire them to the Veroboard as indicated. Next wire the 10 wires from the Veroboard and panel components to the 10 way terminal block, mounted on the back of the case. All wires will need to be about 6 to 9 inches long. Connect the battery leads to the

The transistors TR1 and TR3 and the Zener diode, D1, should now be soldered in position—remember to use a heat shunt on each lead otherwise damage to these devices may result.

TRANSDUCER BOX

Veroboard.

In the prototype, the components were all mounted on the top panel (approximate size 13 x 10 inches) of a Perspex case. Perspex was

chosen for the transducer box because of its excellent weather resistant properties. Any other case material may be used but it will probably need a good coat of paint to protect it.

If a metal case is used some method of insulating the nut, bolt and solder tag connections to RTH1 and the Veroboard sensor must be employed.

The positioning of the components in the prototype is indicated in Fig. 5. This is not critical but there are some important points to watch.

The rotor should not be too close to the vanes such that its rotation will cause the direction vanes to be affected.

The photocell PCC1 should be in an unobstructed position of daylight. Shadows from the rotor or vanes should not be allowed to fall on it at any time.

DIRECTION POTENTIOMETERS

Before assembly of the transducer components, VR5 and VR6 must be modified.

They must be dismantled and their stops removed.

With all potentiometers there is a gap between the two ends (start and finish) of the carbon tracks; this must be bridged with an insulating material otherwise the slider will get stuck in this gap and erroneous measurements will be taken, A suitable material for this is Araldite. Fill the gap, smooth over and leave to set. Ensure, when dry, that the slider can move across this region with minimum pressure.

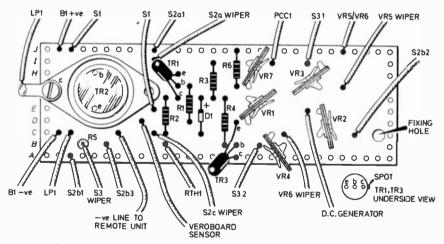
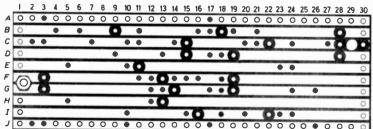
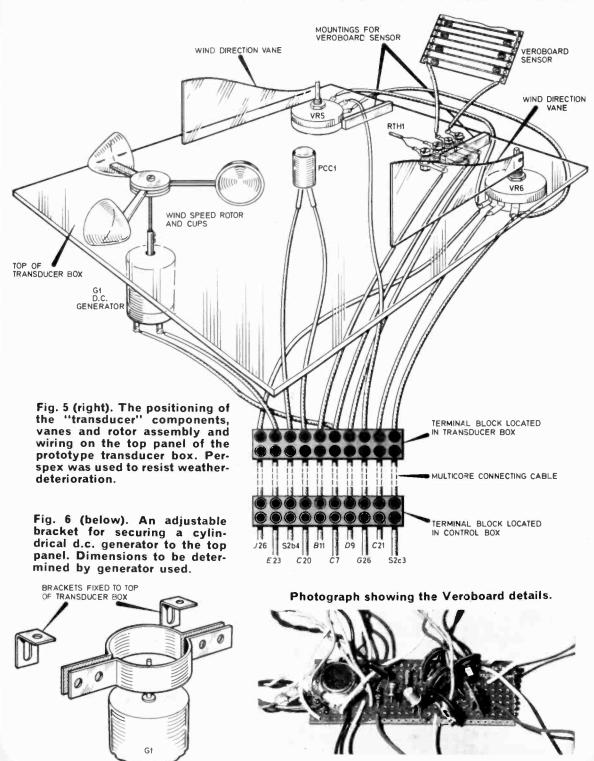


Fig. 3. The layout of the components on the Veroboard for the control box, and the regions of copper strip to be removed from the underside.



weather station



TR2 COLLECTOR PCC1 IN REMOTE UNIT E 10 A3 C 27 86 **[23** 819

Fig. 4. The positions of the components and the wiring on the back of the control box front panel. Note the meter polarity. The layout is not critical and may be changed to suit individual requirements.

ROTOR CUPS AND VANES

The cups and vanes used in the prototype are made of Perspex (see Shop Talk for details).

If desired the cups can be made from suitable plastic egg cups by cutting away the base of the egg cup and filing smooth.

The dimensions of the direction vanes are not critical, except that they should, for efficient and speedy alignment, be triangular in shape, as shown in Fig. 5. These can be bolted to the flat sides of the spindles of VR5 and VR6, suitable holes being drilled in the spindles to accommodate the fixings.

LIGHT CELL AND RAIN SENSOR

The light cell can be mounted in any unshadowed position. In the prototype its position was central on the top panel and raised by means

of a short length of plastic tubing.

A suitable sensor can be made from Veroboard cut and soldered as shown in Fig. 5. approximate size 312 by 3 inches. In the prototype, tapped Perspex 14 x 14 x 2 inches was glued to the base of the sensor and this was screwed in position on the top panel. The thermistor, RTH1, is located under this sensor for convenience and protection.

GENERATOR FIXING

Motors suitable for use as the generator G1 are available in all shapes and sizes. The one used in the prototype is cylindrical, and is attached to the top panel by means of the adjustable bracket shown in Fig. 6. This can be modified to suit different generator shapes.

The rotor may be attached to the generator shaft by fixing a long thin brass bolt through the centre of the rotor system and uniting this to the generator shaft with a spindle coupling.

WIRING

When all the components have been fixed in their final positions on the transducer box, wiring may be carried out in accordance with the wiring diagram of Fig. 5. All wires go to the terminal block located inside the transducer box.

The two boxes are best and easiest united using a long length of multicore cable passed out through a grommet in the transducer box-the length depending on the final position of the two boxes.

The cable must contain at least 10 insulated wires. This type of cable is available from some of our advertisers.

TESTING AND SETTING UP

When the two parts have been connected, the wiring should be thoroughly checked out before switching on.

When you are satisfied that wiring is correct switch on and test each discrete unit individually and calibrate as detailed earlier.

It only remains now for the transducer box to be placed in position in the right direction. This is done with the aid of a magnetic compass.

With S3 in position 1, and S2 turned to the "wind direction" position, point the narrow end of VR5 vane in a due east direction and hold in this position while turning the transducer box until the needle on the meter indicates east (E). Secure the transducer box in this position.

Small plastic cups (such as those used in bottle tops) can be fitted over the wind speed and direction spindles so that no rain or moisture will run down the spindles into the "electronic" parts. A small amount of grease or Vaseline placed around the spindle bases will also provide protection.

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2G374 2G381	20p 22 jp	2N 3572 2N 3605	97 i p 27 i p	40329 40344	80p 27 jp	BCY40 BCY42	50p 15p	B8Y24 B8Y25	15p 15p	NKT452 NKT453	62 p 47 p
2N404	2219	2N 3606	271p	40347	571 p	BCY43	15p	BSY26	174p	N K T6031	7221n
2N696 2N697	209 179	2N3607 2N3702	22 p	40348 40360	52 p 42 p	BCY54 BCY58	22 p	B8Y27 B8Y28	17 p	NKT6131 NKT6741	7 32 jp 7 80 p
2N698	25p	2N3703	105	40361		BCY59	90 ln	B8 729	17 ip 25p	NKT6771	7 20a
2N706 2N705 A	1819	2N3704 2N3705	11p 10p	40362 40370	57 p 88 p	BCY60 BCY70	97 i p 20 p	B8 Y 32 B8 Y 36	25p 25p	NKT713 NKT781	25p
2N708	184p 15p	2N3706	09 p	40406	574 p	BCY71	95 n	BSY37	25p	NKT1041	
2N709 2N718	82 j p 25 p	2N3707 2N3708	11p 07p	40407 40408	E01p	BCY72 BCZ10	174p 274p	B8Y38 B8Y39	22 p	NKT1043	37∤p
2N726	30p	2N3709	00.	40410	82 ip	BCZ11	421p	B8Y40	32 p	NKT1051	9
2N727 2N914	30p	2N3710 2N3711	09 n	40467 A 40468 A	5719 35p	BD116 BD121	\$1-124 85p	B8Y51 B8Y52	32 j p	NKT2032	82 i p
2N916	17 p	2N3715	12p £1.25	40600	57 jp 80p	BD123	62 ł p	BSY53	87∤p		47 p
2N918 2N929	30p 22}p	2N3716 2N3791	£1.30 £2.06	AC 107 AC126	30p 20p	BD124 BD131	60 p 75 p	B8Y54 B8Y56	40p 90p	NKT2033	87 p
2N930	274 p	2N3819	35p	AC127	25p	BD132	85 p	BSY78	4749	NKT6011	1
2N 1090 2N 1091	22 p 22 p	2N3823 2N3854	97 p	AC128 AC154	20p 22ip	BDY10 BDY11	£1-874 £1-624	BSY79 BSY82	45p 52 jp	NKT8011	77}p
2N1131	zop	2N3854A	27 è p	AC176		BDY17	\$1.50	B8Y90	671n		97 p
2N1133 2N1302	25p 17ip	2N3855 2N3855A	27 p 30p	AC187 AC188	62 i p 87 i p	BDY18 BDY19	£1.75	BSY95A BSW41	12 p 42 p	NKT8011	13 #1.12
2N1303	1749	2N3856	30p	ACY17	27 i p	BDY20	\$1-18	B8W70	27 jp	NKT8021	1
2N1304 2N1305	22 jp 22 jp	2N3856 A 2N3858	35p 25p	ACY18	25p 25p	BDY38 BDY50	971p	C111 C424	75p 27ip	NKT8021	92ip
2N1306		2N3858A	. 30p	ACY20	25p	BDY61	\$1.25	C425	88p		92 p
2N1307 2N1308	25p 30p	2N3859 2N3859A	27 p	ACY21 ACY22	25p 20p	BDY62 BP115	£1-00 25p	C426 C428	40p	NKT8021	92 jp
2N1309	80p	2N 3860	80p	ACY28	20e	BF117	474p	C744	871P 80p	NKT8021	l4 I
2N 1507 2N 1613	174p 25p	2N3866 2N3877	£1.50 40p	ACY40 ACY41	20p 25p	BF163 BF167	87 p 18p	D16P1 D16P2	27≩p 40p	NK T8021	92}p
2N1631	35 p 30 p	2N3877A	40m	ACY44 AD140	600	BF173 BF177	19p	DISP3 DISP4	371P 40p		92}p
2N 1632 2N 1638	271p	2 N 3 9 0 0 2 N 3 9 0 0 A	87 i p 40 p	AD149	52 i p 57 i p	BF178	30p 30p	GET102	30p	NKT8021	921p
2N1639 2N1671B	2719 81-00	2N3901 2N3903	97∔p 85p	AD150 AD161	62}p	BF179 BF180	30p 35p	GET113 GET114	20p 20p	OC20 OC22	75p 50p
2N1711	25p	2N3904	85p	AD162	271p	BF181	32 lp	GET118	20p	OC23	60p
2N1889 2N1893	321p	2N3905 2N3906	871p	AP106 AP114	42 p 25 p	BF184 BF185	25p 42 j p	GET119 GET120	20p 52}p	OC24 OC25	60p 50p
2N2147	37 p 82 p	2 N 4058	17ip 10p	AF115	25 p	BF194	174p	GET873	12 p 30p	OC26	27 p
2N2148 2N2160	5719 5719	2N4059 2N4060	10p 12ip	AF116 AF117	25p 25p	BF195 BF196	15p 42}p	GET880 GET887	\$0p 20p	OC28 OC29	621p
2N2193	40p	2N4061	12 è p	AF118	82 i p	BF197	4219	GET889	22 p	OC35	aup
2N2193A 2N2194A	421P	2N4062 2N4244	12 p	AF119 AF124	20p 22 jp	BF198 BF200	42 p	GET896	22 j p 22 j p	OC36 OC41	62 i p 22 i p
2N2217	271 p	2N4285	1749	AF125	209	BF224	52 p	GET897	22 i p	OC42	200
2N2218 2N2219	23p 23p	2N4286 2N4287	17ip	AF126 AF127	20p	BF225 BF237	19p 23p	GET898 MJ400	22ip 11-07i	OC44 OC45	20p 12ip
2 N2220	251	2N4288	174p	AF139	271 m	BF238	93n	MJ420	B1-12#	OC46	15n
2N2221 2N2222	25p 30p	2N 4290 2N 4291	17 p	AF178 AF179	421p 781p	BF244 BFW61	23p 47ip		11.12 i	OC70 OC71	15p
2N2270	47 lp	2N 4292	12 i p	AP180	K91-a	BFX12	221 n	MJ440	96P	OC72	12 p 12 p
2N2297 2N2368	80p 17∦₽	2N4303 2N5027	47 p 52 p	AF181 AF239	42 p	BFX13 BFX29	22 p 30p	MJ480 MJ481	9749 £1.25	OC74 OC75	32 p 22 p
2N2369	174p	2N5028	57 ł p	AF279	474n	BFX30	30p	MJ490	41.00	OC76	2212
2N2369A 2N2410	174P	2N5029 2N5030	47 p	AF280 AF211	82 p	BFX42 BFX44	87 p	MJ491 4 MJ1800 4	11-87	OC77 OC81	30p 20p
2N2483 2N2484	2719	2N5172	12ip 52ip	A8Y26 A8Y27	25p	BFX68 BFX84	67 p 25 p	MJE340 MJE520	621p	OC81D OC83	221 p 25 p
2N 2539	22 ip	2N5174 2N5175	52 p	ASY 28	871p	BFX85	32 į p	MJE521	78p	OC84	25p
2N2540 2N2613	22ip 35p	2N5176 2N5232A	45p 80p	ASY29 ASY36	27 i p 85 p	BFX86 BFX87	26p 27∤p	MPF102 MPF108	48 p 87 p	OC139 OC140	38ip 38ip
2N2614	30p	2N5245	45p	ASY50	96-	BFX88	96	MPF104	87 P	OC170	309
2N2646 2N2696	52 i p	2N5246 2N5249	421p	A8Y51 A8Y54	22 j p 25 p	BFX89 BFX93	68ip	MPF105 MPR3638	37ip	OC171 OC200	80p 40p
2N2711	200	2N5265	48-25	A8Y86	32 ł p	BFY10	38 ł p	NKT001:	3 474 P	OC201	60p
2N2712 2N2713	25p	2N5266 2N5267	£2-75 £2-62}	AU103 ABZ21	42 i p	BFY11 BFY17	421p	NKT124 NKT125	42 p 27 p	OC202 OC203	75p 42 jp
2N2714	271p 30p	2N5305	871p	BC107	TOD	BFY18		NKT126	271P	OC204	424 p
2N2865 2N2904	62 i p	2N5306 2N5307	40p 87 jp	BC108 BC109	10p 10p	BFY19 BFY20	32ip	NKT128 NKT135	27 p 27 p	OC205 OC207	90p 75p
2N2904A	891p	2N5308	27 lp	BC113	15p	BFY21	421p	NKT137	32 ip 80p	OCP71	42 ip 50 p
2N2905 2N2905A	40p	2N5309 2N5310	621p	BC115 BC116A	15 p 15 p	BFY24 BFY25	45p 25p	NKT210 NKT211	80p 80p	ORP12 ORP61	50p 50p
2N2906 2N2906 A	25 p	2N5354	27 p	BC118 BC121	10a	BFY26 BFY29	20n	NKT212	80p	P346A	221n
2N2907	27ip 30p	2N5355 2N5356	27 ip 32 ip	BC122	20p 20p	BFY30	50p 50p	NKT213 NKT214	30p 22 p	T1834 T1843	82 p
2N2923 2N2924	15p 15p	2N5365 2N5366	4719 8819	BC125 BC126	20p 20p	BFY41 BFY43	60m	NKT215 NKT216	22 i p	T1844 T1845	10p 10p
2N 2924 2N 2925	15p	2N5367	57 p	BC140	37 p	BFY50	62ip 23p 20p	NKT217	87 P 42 P	T1846	110
2N2926 Green	14p	2N5457 2B005	37 p 75 p	BC147 BC148	10p 10p	BFY51 BFY52	20p	NKT219 NKT223	80p	T1847 T1848	11 n
Yellow	12 9	28020	\$2-00	BC149	12p	BFY53	1716	NKT224	27 i p 25 p	T1849	12 p 12 p 17 p
Orange 2N 3011	12ip	28102 28103	50p 25p	BC152 BC157	17+p 20p	BFY56A BFY75	571p 80p	NKT225 NKT229	22 j p 30 p	T1850 T1851	17 p
2N3014	30p 32ip 18p	28104	25a	BC158	11p	BFY76	42 i p	NKT237	35p	T1852	12 p 12 p
2N2053 2N3054	18p 46p	28501 28502	321p 35p	BC159 BC160	120	BFY77 BFY90	574p	NKT238 NKT240	25p 27 i p	T1853 T1860	22 p 22 p
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2N3133 2N3134	80p	3N83 3N128	70s	BC168B BC168C	10p 11p	BFW59 BFW60	25p 25p	NKT242 NKT243	621p	T1862 T1P29A	87↓p 50p
2N 31 35	25 p	3N 140	77†p	BC169B	11p	BPX 25	\$1.85	NKT244	174P	TIP29A TIP30A	80p
2N3136 2N3390	25p 25p	3N141 3N142	78 i p 55 p	BC169C BC170	12p 13ip	BPX29 BPY10	\$1.80 £1.45	NKT245 NKT261	20p 20p	TIP31A TIP32A	63 ≟p 75p
2N3391	20a	3N 143	67 p	BC171	15p	BRY39	871p	NKT262	80p	TIPSSA	
2N3391A 2N3392	30p 17ip	3N 152 R.C.A.	87 p	BC172 BC175	15p 22 jp	B8X19 B8X20	174p	NKT264 NKT271	20p 20p	TIP34A	-02 p 42-05
2N 3393	15p	40050	550	BC182	10p	B6X21	87 p	NKT271 NKT272	20p	TIP35A	42-90
2N3394 2N3402	15p 23 j p	40251 40309	38 p	BC183 BC184	09p 11p	B8X26 B8X27	45p	NKT274 NKT275	20p 20p	TIPSGA	99.02
2N3403	28 j p	40310	407	BC212L	189	B8X28	32 j p	NKT281	87 ł p		
P	ost &	Matchine	Jp per	e (audio	transie	tors only	0mmon	wealth (A	Air) 65 Dair	p (MIN.)	
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	1-11	12-24		1-11	18-24		1-11	12-24
	₫p	£p.		£p	£p		åp.	£p
BN7400	0.20	0.18	SN7433	0.80	0.75	8N7472	0.32	0.30
BN7401	0.20	0.18	8N7437	0.64	0.06	BN7473	0.48	0.41
8N7402	0.20	0.18	8N7438	0.64	0-60	8N7474	0.43	0.41
8N7403	0.20	0.18	8N7440	0.28	0.21	8N7475	0.45	0.44
8N7405	0.20	0.18	SN744LAN	0.87	0.88	BN7476	0.45	0.44
BN7406	0.80	0-75	BN7442	0.85	0.81	8N7480	0.70	0.65
8N7407	0.80	0.75	BN7443	2.86	2.70	8N7481	1.40	1.88
BN7408	0.20	0.18	BN7444	2.86	2.70	8N7482	0.87	0.88
BN7409	0-20	0.18	8N7445	2.50	2-40	BN 7483	0.87	
BN7410	0.20	0.18	BN7446	1.00	0.95	BN7484	2.00	1.85
8N7411	0.23	0.21	BN7447	1.00	0.95	BN7485	3-68	3-40
SN7412	0.48	0.46	BN7448	1.00	0.95	8N7486	0.88	0.30
SN7413	0.40	0.38	BN7449	1.00	0.95	BN7490	0.87	0.84
SN7420	0.20	0.18	BN7450	0.20	0.18	8N7491AN	1-21	1.10
8N7423	0.51	0-47	SN7451	0.20	0.18	8N7492	0.87	0.84
8N7427	0.48	0-45	BN7453	0-20	0.18	BN7493	0.87	0.84
BN7428	0.80	0.75	8N7454	0.20	0.18	BN7494	0.87	0.84
BN7430	0.23	0.15	BN7460	0.20	0.18	8N7495	0.87	0.84
8N7432	0.48	0.42	8N7470	0.40	0.38	BN7496	0.87	0.84
	S	UB-I	MIN ELI	ECT	ROL	YTIC		_
range axi	al lead						61	each
Values: i	(uF/V):	0.64/	64: 1/40: 1	6/25:	2.5/16	3: 2:2/63:	4/10:	4/40:

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		SILI	CON	REC	CTIFI	ERS		
PIV	50	100	200	400	600	800	1000	1200
1.4	8p	92	10p	11p	18p	159	20p	_
3 A	15p	17p	20p	22p	25p	27p	80p	35p
6A			25p	80p	32 p	85p	_	_
10 A	80p	35p	40p	47p	56 p	86p	75p	_
15 A	36p	45p	48p	55 p	65p	75 p	87p	_
35 A	70p	80p	90p	\$1.00	#1-40	\$1.70	42.75	_
I amp a	nd 8 ams	are pla	stic enc	apsulati	ion.			

		DIODI	ES &	RECT	IFIER	lS	
IN34A	10p	AA119	7p	BAX16	12i n	F8T3/4	2319
1N914	7p	AA129	15p	BAY18	174p	OA5	17p
1N916	7p	AAZ13	12p	BAY31	7p	OAIO	20p
IN4007	20 p	AA715	129	BAY38	25p	OA9	10p
1844	7p	AAZ17	10p	BY100	15p	OA47	8p
18113	15p	BA100	15p	BY103	22p	OA70	7p
18120	12p	BA102	25 p	BY122	474p	OA78	10p
18121	14p	BA110	25p	BY124	15p	OA79	7p
19130	8p	BA114	15p	BY126	15p	OAB1	85
16131	10p	BA115	70	BY127	17p	0A85	10p
18132	12p	BA141	7p 17p	BY164	579	OA90	7p
18920	7p	BA142	17p	BYX10	22 p	OA91	79
18922	8p	BA144	12p	BYZ10	25p	OA95	7p
18923	12p	BA145	17p	BYZ11	32p	OA200	70
18940	5p	BA154	12p	BYZ12	80p	OA202	10p
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v	ΕF	OB	DAF	RD.	
				0.15	0-1
				Matrix	Matrix
		3 in		17p	28p
		5ln		25 p	25p
31		3 in		25 p	25p
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THEY ADE EIR MARK **104 Ohm**

By J. E. Gregory

Photograph: Radio Times Hulton Picture Library.

HIS month we move to Germany, to hear about the man who gave his name to the unit of resistance, Georg Simon Ohm, and his now famous law (see Table 1).

The Ohm family lived in Erlangen, South Germany with their two sons Georg and Martin. Ohm senior was in business in a modest way as a locksmith, his wife complimented the family income by taking lodgers and it was one of these who was to influence the eldest son Georg, born March

16, 1787, in his early years.
It was expected that Georg being the eldest would take over the family business, but having been inspired by the student lodger to study he entered the local university. He was a competent pupil, and a natural born teacher, and on completing his course, accepted in 1817, at the age of 30, the post of professor of mathematics at the Jesuits college in Cologne.

OHM'S LAW

During this time he set his mind to many problems contained in mathematics, physics, and the new power electricity. It was at Cologne he learnt how the "Flux of Heat" in a metal bar is directly proportional to the difference of temperature between its ends, and then, influenced by the journals of Schweigger Poggendorf the German physicists which had appeared in 1825. Ohm sought to discover what laws regulated the flow of electricity in a conductor.

His results were published in Berlin in 1827 in a pamphlet with the title "Die Galvanishe Kette Mathematisch Bearbeitet" which roughly translated means "The Mathematical Works of the Gal-

The oldest and one of the most important electrical "laws" was formulated by Georg Ohm, and the unit of resistance—the ohm—is named after him. This is used to measure the value of the resistor, the most commonly used of all electronic components.

TABLE I THE OHM (Ω)

Ohm's law illustrates that the resistance of a conductor is equal to the voltage across it divided by the current flowing through it. When the resistance is in ohms the voltage is in volts, and the current in amps.

The Omega sign (Ω) used for the ohm was suggested by Sir William Henry Preece the British electrical engineer and wireless telegraphy pioneer, whilst lecturing to Indian telegraph service cadets at the Hartley College, Southampton in 1867.

vanic Chain". The theory Ohm proposed set out that the electric current is a conductor is directly proportional to the difference of potentials between its ends.

The publication of the pamphlet caused a major upset in German scientific circles, people said Ohm was mad, and his law absurd, he was even forced to resign his post at Cologne.

For six years he was an outcast doing very little scientific work then in 1833 he obtained a post at Nuremberg polytechnic.

RECOGNITION

Gradually, however, the pamphlet began to get a wider circulation and his law was being quoted. Then, quite suddenly in 1841, came the first official recognition, Ohm was awarded the Copley medal of the Royal Society and made a foreign member of the society.

This recognition from Britain, and one of Europe's oldest scientific societies inspired Ohm to return to his work. He published a number of papers on mathematics and a memoir on interference in uniaxial crystals.

Ohm's apparatus used in his work with his law on the Galvanic circuit.

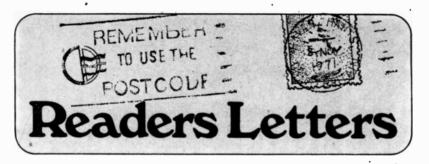
In 1849 he was appointed to the important post of conservator at the physical collection at Munich, and in 1852 professor of experimental physics in the high school at Munich, he started work on a text book of physics which was published in 1854 just before he died of apoplexy brought on by overwork, on July 7.

Thus at 67 died the man who in the face of great opposition took the first steps towards the formulation of the laws of the electric current.



Photograph: Crown copyright Science Museum.

Everyday Electronics, August 1972



Resistor Symbols

I have read every issue of EVERYDAY ELECTRONICS and I have only just noticed, in reading your Guide To Circuit Symbols, that you have been using the old symbol for resistors (a zig-zag line). I use many drawings of electronic circuits in my work and all up to date drawings now use the B.S. symbol (an oblong box) for a resistor. Whether this is done on purpose, to make it plain for the beginner I do not know, but I just thought I would point it out to you.

K. C. Vicars, Portsmouth.

The symbol we use is given by B.S. as an alternative; see this month's editorial for comment.

That 15p Effort

As one who was brought up on thermionic valves and was introduced to electronics largely through your parent magazine, P.E., I should like to add to the congratulations you have received on the success of E.E.

Although I have experimented with radio for over 25 years, it is thanks to your magazine that I am beginning to feel at home with solid state devices.

Unfortunately, some of your ungrateful critics have yet to learn to walk before they can run, and it is hardly the function of E.E. to provide detailed modifications and servicing for commercial devices. Lack of effort, indeed, Mr. Alexander! Some people do not realise how much effort must go into making E.E. the best value on the bookstall—and all for 15p a month!

Mr. D. B. Lyall, Cheltenham.

Modification

It may be of interest to you that the circuit given in the May issue of your magazine for a Bee

Counter was used as the basis of a paper counting machine for our school magazine. I had been planning to make such a device using a photocell, one transistor amplifier and a relay, thus operating a magnetic counter. Your circuit was a welcome alternative, being less bulky and requiring only one power source.

I built the circuit as described and arranged that the light beam should be broken by the sheet of paper falling vertically onto a chute which guided it into an output tray. In practice this arrangement failed to operate because of the small change in light level brought about by the sheet of paper.

This was due to two factors: firstly the paper is rather thin, and secondly, the interference from the room lighting which was not interrupted by the paper. Thus the circuit had to be made more sensitive to smaller changes in light level. To do this I introduced a further stage of amplification using a second OC72 and its two associated resistors. With this slight modification a very successful paper counting device was constructed.

I continue to be impressed with the variety of useful projects that you publish and wish you all success in future publications.

> R. Anderson, Leeds.

"B" Class

We followed with great interest your article on the electronic Bee Counter. However, we thought your entrance and exit labels a little strange because we don't know a lot of bees capable of reading and writing.

Of course, we have heard that the educational system has been greatly modified in the United Kingdom of late, presumably the "A" class is now reserved for people and the, dare we say it, "B" class, for insects.

Five unreadable signatures, No address.

We must admit that we do not know of any educated bees, but if you know some—as you infer—we are sure many bee keepers would be interested. Perhaps you could tell us how the bees can read the signs when they are inside the hive?

To be more serious, if you look at page 380 you will find that the caption to the photograph refers to the labels, thus constructors (not bees) are made fully aware of which aperture is which.

Transistor Holders

I have a query to put to you. When I was at school, 15 years ago, transistors had just come into general use, and I had many knowledgeable friends who tried to introduce me to the hobby of electronics; unfortunately I never quite got to grips with the practical side and so never got going. This is why your magazine is so helpful to me, for example I now understand for the first time in my life how to solder a good joint.

My query is this; these friends of mine at school were unanimous in recommending the use of transistor-holders, to avoid the dangers of soldering direct to the leads; as you do not seem to mention these, I am wondering whether there is some good reason that their use has been discontinued.

N. E. Goller, London, N.W.3.

Transistor holders are still available and can be used on almost any project if desired. However if some care is taken when soldering, the transistors should not be damaged in any way. Silicon transistors can of course stand much more heat than the original germanium types.

Rain Wanted

How Jucky you are in the UK., you get your June issue in June. Here, the April issue has just hit the bookstalls. It's worth waiting for though. For years now, I've been wading through electronics magazines but the penny didn't drop until your *Teach-In* penetrated my grey matter.

Sinclair Project 605 Amplifier Project 605 the new simple way to assemble Sinclair high fidelity modules rik connector unit



For several years now you have been able to assemble your own high fidelity system to world beating standards using Sinclair modules. We have progressively improved these technically but hitherto the method of assembly at your end has remained the same - there has been no alternative to a soldering iron. Now for those who prefer not to solder, there is an alternative - Project 605.

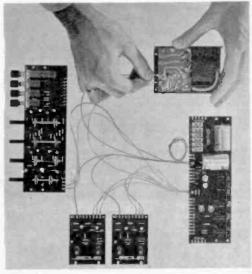
In one neat package you can now obtain the four basic Project 60 modules plus a fifth completely new one - Masterlink - which contains all the input sockets and output components you previously bought separately. Also in the Project 605 pack are all the inter-connecting leads, cut to length and fitted at each end with plugs which clip straight onto the modules, eliminating soldering completely. The pack contains everything you need to build a complete 3C watt stereo amplifier together with a clear well illustrated Instruction Book. All you have to do is to arrange your modules in the plinth or case of your choice and then clip them together - the work of a few minutes.

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I agree with your reader who asked for more mathematical tolerance. It became necessary for me to take a refresher course in Algebra so that I could follow your development of equations in the resistors in parallel bit.

Perhaps you are not so lucky there in the old country, we don't have that awful problem of where to go and what to buy when stocking up for one of your projects. It's a case of "shop a" or "shop b" or write off to the Eastern States, some 3000 miles away, and wait several weeks.

Your Rain Warning Alarm appealed to me and after wrecking it with reverse battery connections as per your erroneous instructions, I eventually got it problem Just one working. though. Its been installed on the roof for about 2 months nowbut it hasn't rained!

How about a pull out data page suitable for clipping into a workshop manual. Such information as colour codes, transistor base connections, fundamental circuits like the diode pump, hints, tips and short cuts gathered together in one handy cover would be a boon for the likes of me who hit a problem and have to go searching through piles of magazines to find a clue to what could be wrong.

> Ray Foster, Western Australia.

To be correct our June issue is on sale during May in this country. By the time you read this you should have the Constructors Companion, presented free with the May issue—we hope this will meet your needs.

Prices

As many other readers have said, your magazine is a good one and has cleared up a few mystifying points. But now I would like to clear up a point myself, just in case someone got the wrong idea about New Zealand from Mr. J. Koppard's letter in E.E. April 1972. New Zealand is not a pin prick on the map down under where people walk round in grass skirts, but a modern thriving country.

The price of the OC71 is something of a mystery to me. Here in Auckland an average OC71 cost 70c or 35 English pence, not as cheap as in England but still a far cry from £3. I do not know

the price of an OC71 in Wellington but since Wellington is the capital of N.Z., I doubt if the price would be higher.

My guess is that the price or type of article was confused somewhere along the line, but if this is not the case then Mr. J. Koppard paid an extra £2.65p to a very low dealer.

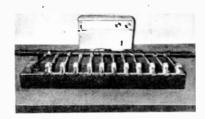
E. Van Dyke, New Zealand.

Gas Station!

Sensor should eat his words. I have enclosed a photograph of a gas operated radio which the East Midlands Gas Board uses for demonstration. What is more, the device even works on natural

G. A. Bolton Shepshed, Leicestershire.

Sensor did not say that it could not be done, merely that "there are limits to what can be done by gas". However we thank you for the inclusion of the photograph, which we show below.



Electronic Course

During the last two years, I have been running a course in basic electronics for the student with a non-vocational background. This course has a great deal in common with the "Teach-In" articles that you have been publishing since November 1971, and it has been suggested to me

readers that your may be interested

The course is based on attendance at the college for one evening per week (7.00-9.00 p.m.) for 24 weeks and runs from September to Easter. There is a strong practical background to the instruction, and demonstrations and experiments are conducted almost every night of the course.

I hope you find this suggestion of interest.

> H. May, Head of Electrical Trades Department, Gateshead Technical College, Durham Road, Gateshead.

Thanks

Further to my letter in the June issue, I now know the secrets of Snap Sequence, Indicator and Home Sentinel, thanks to an unknown reader who rang me and sent me the first issue.

I would like to refund the postage to him, and also return his magazine (if he would like it back) if he will let me know his address. I am afraid I put the phone down before I realised I didn't know who it was.

R. Brown, Burton-on-the-Wold.

Component Export

We would be obliged if you could put us in touch with some firms in the UK who would be interested in the export of electronic components and gadgets to an electronic agency here.

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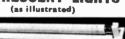
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В	C183L C184L	10n	MPSU56	50p	2N697	15n	2N5459 40p
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