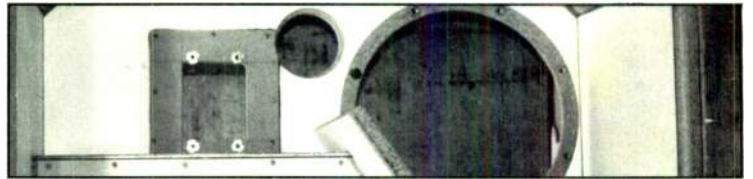




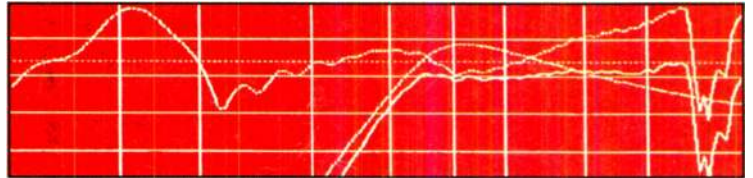
Speaker Builder

THE LOUDSPEAKER JOURNAL

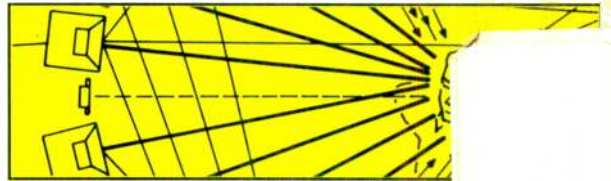
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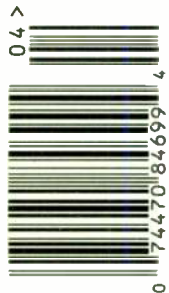
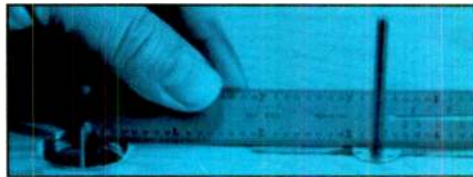
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World Radio History

Good News

■ AUDIO CABLE

An audio cable engineered and optimized specifically for biwired speakers has been introduced by MIT. T2 Bi-Wire is based on the patented MIT Terminator™ technologies, which are designed to provide proper bass reproduction and midrange clarity with an efficient interface. A more recent development is the Multi-Bandwidth™ Terminator, in which separate networks are optimized for high and low frequencies. Music Interface Technologies, Auburn, CA 95602, (916) 888-0394, FAX (916) 888-0783.

Reader Service #102

⇒ DRIVER PROTECTION

Speaker-Knife is a high-speed transient/overload speaker-driver protector. Once triggered, the device either cuts out the audio and stays disconnected until the fault is corrected or instantly resets. Drivers and crossovers are protected from over-excitation, blown voice coils, and constant DC. The absence of batteries means there is no robbing of usable audio power or frequency response/dynamic range loss. All hardware is included. Dynastar Electronics, 926 E. Birch Ave., Fresno, CA 93720, (209) 435-0270.

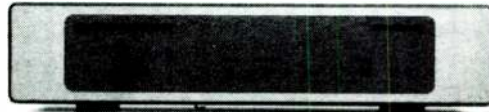
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⇒ DIGITAL PROCESSOR

TransDAC is the assembled version of the Assemblage DAC-1 DIY kit (available through The Parts Connection). The small chassis (9.5" x 2" x 7") houses a potted toroidal power transformer, low-ESR power supply capacitors, Crystal CS8412 input receiver, NPC 5813 digital filter, and two Burr-Brown PCM1702 20-bit DACs. The output stage is a direct-coupled, Class A design, utilizing high-speed Analog Devices AD844 and AD847 op amps, film and foil polypropylene capacitors, and 1% metal film resistors. A special introductory promotion pairs the TransDAC with the UltraJitterbug jitter-rejection device. Sonic Frontiers, Inc., 2790 Brighton Rd., Oakville, ON, Canada L6H 5T4, (905) 829-3838, FAX (905) 829-3033.

Reader Service #105



■ MEASUREMENT SOFTWARE

Liberty Audiosuite is a new measurement and analysis software package, utilizing multimedia soundcards based on the ADSP2115 DSP. The software is capable of measuring frequency and time-domain responses, impedances, spectra, distortion, anechoic behavior, room acoustics, and other characteristics of audio frequency electronics and loudspeakers. A choice of 14 sample rates from 5.5kHz to 48kHz is provided, with MLS/FFT and sine wave capability to 16,384 data points. Liberty Instruments, Inc., PO Box 1454, West Chester, OH 45071, phone/FAX (513) 755-0252.

Reader Service #101

■ WINDOWS FOR SOFTWARE

A Windows version of the Listening Room is now available. The program helps locate speaker and listener placements to minimize the deleterious effects of standing waves and boundary reflections. Besides a graphical interface, the new version features subwoofer placement, electrostatics, surround-sound speakers, openings and drapes, consideration of enclosure size and speaker placement on baffles, and accurate scaling of room diagrams. Windows 3.1 and a mouse are required. Sitting Duck Software, PO Box 130, Veneta, OR 97487, (503) 935-3982.

Reader Service #104

■ SPEAKER COMPONENTS

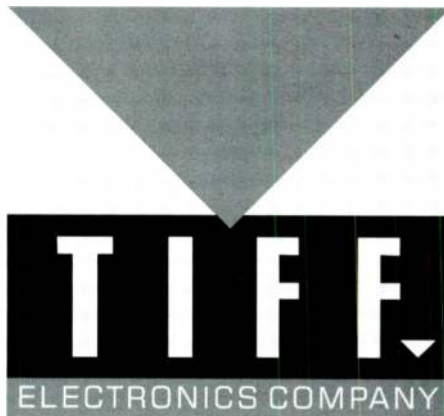
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Reader Service #106

Speaker Builder (US ISSN 0199-7929) is published every six weeks (eight times a year), at \$32 per year, \$58 for two years; Canada add \$8 per year; overseas rates \$50 one year, \$90 two years; by Audio Amateur Publications, Inc., Edward T. Dell, Jr., President, at 305 Union Street, PO Box 494, Peterborough, NH 03458-0494. Second-class postage paid at Peterborough, NH and an additional mailing office.

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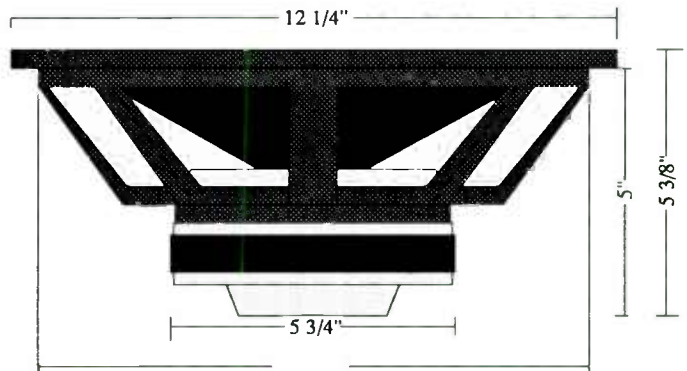
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Reader Service #72

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EFFECTIVE CONE AREA	507 cm ²
VOICE COIL RESISTANCE	3.4 ohm
VOICE COIL INDUCTANCE	1.2 mH
VOICE COIL DIAMETER	50 mm
VOICE COIL HEIGHT	34 mm
AIR GAP HEIGHT	8 mm
XMAX	13 mm
FREE AIR RESONANCE	19 Hz
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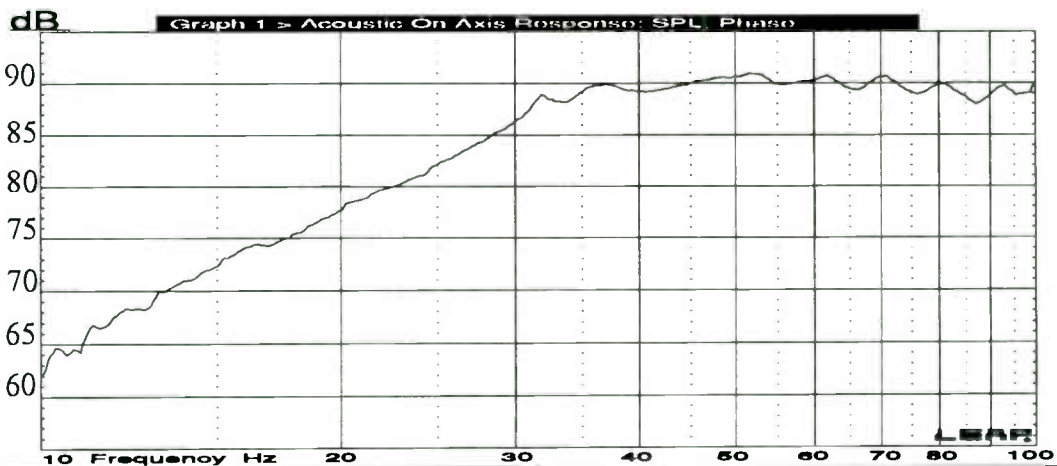
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Editorial

The Computer Question

By now we are all weary, not to say jaded, when the discussion turns to speedy technological change. But such shifts in the way we do things need some perspective if we are to fulfill one of the most basic of philosophical precepts. Socrates said it well: "The unexamined life is not worth living."

My subject is computers. Whatever you think or have heard, or however you may view the computer revolution, step back with me for a few moments and consider some of the main issues.

The company which produces the magazine you hold in your hands has been doing its work and been managed by some form of computer technology since 1982. On June 1, we retired our first system, put in place 13 years ago and only now being decommissioned because of ongoing integration of the way we do business. Today our office houses 23 machines which handle circulation fulfillment, advertising sales management, financial controls, editing manuscripts, page production, and sales and shipping.

Eventually, we hope to offer online services for authors and readers alike, and at least one totally electronic periodical is in the works. You will have noticed that more and more of what appears in *Speaker Builder* is related in one or more ways to the computer as a multipurpose tool. It offers design capability for boxes, crossovers, zobel, not to mention fully integrated, multidriver systems. It allows you to first build a system on the computer, before building it in real time and space, as Vance Dickason's new *Loudspeaker Recipes*¹ illustrates so eloquently.

The computer's measurement capabilities are already pervasive and growing more refined with each passing day. Indeed, a well-equipped moderately powerful machine can fulfill most of the normal electronic test bench functions.

Unlike most technology, however, the cost of the computer has been dropping steadily since its introduction to wide use in the early 1980s. Our first IBM clone was an XT with 1M of memory, a mono monitor, two 5¼" 360K disk drives, and a 20M hard disk, all for \$1750. Today you can purchase an 80486 DX100 machine with 8M of memory, color monitor, 560M hard drive, and 1.2 and 1.4M floppy drives for that same price tag.

If you have put off doing anything about a computer for your speaker building activities, it may be because the whole matter seems so confusing and the vendors so unfamiliar that you just haven't made a start. Since this magazine is based on the do-it-yourself premise, I suggest you build your own. If that seems outrageous, it may be that you need a reliable, comprehensive introduction to the whole topic. I do make the suggestion without any reservations or trepidation, however. If you can use a screwdriver and read instructions, you can build your own machine, and keep it upgraded as the technology develops.

McGraw-Hill's *Windcrest* division has been publishing a series of guidebooks about computer hardware under the title *Build Your Own...PC*.² Aubrey Pilgrim, the series' author, has produced volumes on building all models from the lowly 8088, the IBM XT, to his latest which explores the Pentium processor-based devices. I might explain that there have so far been four successors to IBM's first design. These are all prefixed by the 80... designation. We've had the Advanced Technology (AT) step up, the 286, followed by the 386 and the 486. The latter number is a highly sophisticated device containing a math co-processor, which in the earlier 386 had been a separate chip.

All the while prices of these central processor units (CPUs) dropped steadily, plummeted in some instances, from their introductory prices. The increasing sophistication of the CPUs was accompanied by increases in processing speed. From the 4.7MHz per second XTs to the most recent Pentiums (aka 586) at speeds of 120MHz, we have seen a dazzling increase in the power of these smaller and smaller machines.

Pilgrim's latest volume, *Build Your Own Pentium Processor PC*, may seem to a beginner like starting at the top. In one sense that is so, but the book is an absolutely comprehensive first look at the whole machine, and neither its processor nor its speeds will be any bar to your understanding of how the whole IBM PC series of computers works and how to put one together—regardless of whether you intend to build a 386, a 486, or one of the new Pentiums. The basic elements of all the machines are the same.

There *are* variations, of course. You'll have choices of bus formats, monitors, disk drive formats, memory configurations, and much else. The basic hardware is absurdly easy to assemble, however. The box with power supply, the motherboard, memory, floppy disk(s), disk interface and ports board, monitor interface, hard disk, keyboard, and monitor can all be put together with moderate care using a Philips screwdriver. But Pilgrim takes you on a very comprehensive, easy-to-understand tour of IBM computer country that is the best overview I have seen for beginners. The Pilgrim book is also a handy manual for an experienced user's full understanding of the machine's basic operation.

Computers encompass a very large array of diverse information, and there are seemingly endless things which may be learned about them. A multitude of disciplines exist within the computing world, any one of which by itself can become a full-time avocation. After you have built your computer and installed an operating system, the software choices are a new, and larger, adventure.

You will be seeing more about computers in these pages in upcoming issues. The computer is also a music and data processing instrument. It has the developing capability to record and edit digitally encoded music, and it can be a handmaiden to composers, capturing and printing musical notation, as well as the sound itself.

Inevitably the computer will affect loudspeakers at all levels from materials research right to performance in your listening space. I suspect it will redefine the way we deal with issues of quality and performance. It is already pouring out mountains of new information about all the elements which comprise the loudspeaker interface. If you have been hanging back, it is time to step up across the new threshold and start finding your way around a new, exciting neighborhood.—E.T.D.

REFERENCES

1. Vance Dickason, *Loudspeaker Recipes*, Audio Amateur Press, 1994, ISBN 1-882580-04-4.
2. Aubrey Pilgrim, *Build Your Own Pentium Processor PC*, Windcrest/McGraw-Hill, 1994, ISBN 0-07-050164-5, \$19.95.

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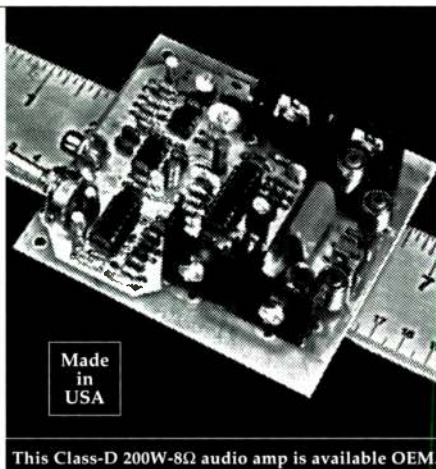
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The peculiar evil of silencing the expression of an opinion is, that it is robbing the human race; posterity as well as the existing generation; those who dissent from the opinion, still more than those who hold it.

∞ JOHN STUART MILL

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Subscription Inquiries

A one year subscription to *Speaker Builder* is \$32. Canada please add \$8. Overseas rate is \$50 per year.

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About This Issue

Speaker Builder has always stressed the importance of reader feedback. Our lead article in this issue is a direct response to your interest in a published project and proof that your comments and suggestions do make a difference. Spurred by a reader request, contributing editor John Cockcroft reveals several woofer solutions (with a surprising orientation) for his Simpline speaker system ("The Simpline Sidewinder Woofer," p. 8). In keeping with the Simpline tradition, this project features simplicity with pleasing results.

If room effects plague your sound system, we offer two solutions. First, Bill Waslo demonstrates his analytical prowess with an array of drivers to deal with the effect of room acoustics on sound quality ("Focused Arrays: Minimizing Room Effects," p. 10). His proposal is an untested, but feasible, way to increase direct sound while reducing room reflections.

Second, contributing editor G.R. Koonce has developed programs for the IMP Audio Analyzer to help you establish the proper test setup and allow IMP files to be used by other programs to achieve a reflection-reduced environment ("Extending IMP: A Program Set," p. 24).

Ron McGhie offers a flexible, four-way vented speaker design that addresses the lack of dynamic range in home systems. His construction project results in a high-performance system in a modest box design ("A Flexible Four-Way System," p. 14).

In his latest installment of Wood World ("Shaped Hole Techniques," p. 40), Bob Wayland presents simple methods to produce professional-looking circular holes to mount a speaker, without the necessity of using heavy-duty equipment.

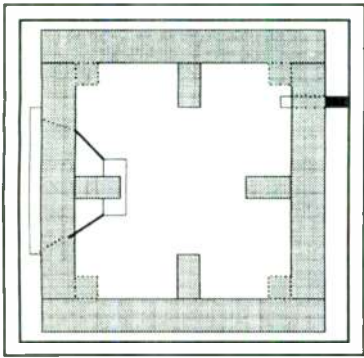
Finally, sound man Dan Ferguson reviews Sheffield Lab's new test CD, appropriately titled *My Disc*, to help you evaluate and analyze your system ("CD Review," p. 48). This useful tool features over 80 technical tracks and sound effects.

Speaker Builder

THE LOUDSPEAKER JOURNAL

VOLUME 16 NUMBER 4

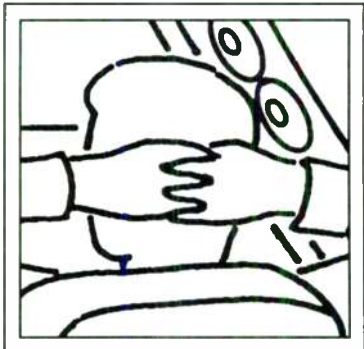
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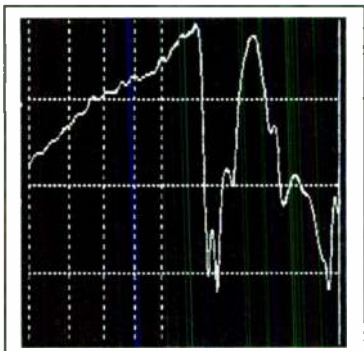
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THE SIMPLINE SIDEWINDER WOOFER

By John Cockroft

Of all of the speaker systems I have designed, the Simplines have made me the happiest and perhaps surprised me the most. They ask so little and give so much (sort of like the Little-Engine-That-Could).

From time to time I have considered using the Simplines as satellites, but for some reason never got around to designing a specific woofer for them. A letter from reader Harry Campbell renewed my interest in the matter.

THE SQUATLINE

I currently lack the facilities for construction, so I can't really design and test a woofer from scratch. But I have been able to do something. I dragged out an old speaker system which I call the Squatline. It was dutifully standing against a wall in the bedroom, trying hard to prevent dust from settling to the floor.

I sometimes thought of this speaker as a 2' long Shortline and at other times as a Microline for an 8" speaker. I disconnected the tweeter and the crossover, and hooked it up to the Simplines through an active crossover consisting of a couple of Marchand XM 1 PC boards and a homebrew power supply. At various times the crossover point was 125Hz and 200Hz.

After a lot of fiddling around, I eventually found what was seemingly the best location for the woofer cone in relation to the room. This turned out to be as close as possible to the junction between the floor and the rear wall, a position originally discovered and used by Roy Allison. In order to achieve this location, the Squatline (which, like the Simplines, has its woofer on its top face) had to lie on the floor with its back against the wall like some boxy house cat.

In this position, when I finally got the levels between the woofer and the satellites set correctly, I spent some time listening to different things: organ, heavy orchestral passages, things with a lot of transients, some swing and Dixieland, and vocals. The results were surprisingly good. The transition between the speakers couldn't be detected; everything seemed to float above the pile of

dead bones leaning up against the wall, totally inert and unconnected with the music. I was really enjoying this.

In the midst of this delicious pleasure, a black thought intruded. The woofer in the Squatline was an old Peerless model not currently being manufactured. *Ow*—that spoiled my day! I later realized I had an 8" Radio Shack 40-1024 woofer (their current 8" poly coned unit) in another system, which I had used for a while in the Squatline. At the time it had sounded OK. I located it, reinstalled it, and reset the balance. When I heard the sound, I was greatly relieved.

A SIDEWINDER IS BORN

Something would have to be done about the enclosure, though. It couldn't just lie around the floor in such an unhousebroken manner. The thing to do was to keep the same dimensions (because the music still sounded fine), but to build it upside down with the speaker at the bottom, on the side panel (see "Construction Details"). The Sipline Sidewinder—*why not?*

My first thought was that the active crossover wasn't exactly in keeping with the concept of simplicity. So I designed a passive crossover and breadboarded it on a piece of cardboard photo mount stock. It was a second-order filter with a Q of 0.5, and it crossed over at about 159Hz using 0.1µF capacitors. While it sounded beautiful, it wouldn't play very loud. My system consisted of a Carver CD player hooked up directly to an amplifier. There wasn't enough get up and go in the CD player to push through the passive filter.

After sitting around for a while with a pessimistic profile, I went to the closet and located my old Hafler

100 preamp. The 20dB gain in the preamp made the difference; it was playing loudly enough to satisfy me (and probably to annoy my neighbors). This system doesn't have to be *really* loud to sound good. Just moderately loud, or even a bit less, and you get a tremendous subjective feeling of reality.

I recently redid the crossover in an effort to make it as low impedance as possible (Fig. 1). It helped—the system will now play louder than I need.

FINDING RESISTOR VALUES

The crossover is designed so the amplifier's input impedance serves as the final resistor of the circuit's high-pass section. I know of no amplifier with the correct impedance. (And if I did, I don't have it.) The purpose of resistor(s) Rx in the circuit is to fool the crossover into thinking the impedance is correct. It must be of a value that, when paralleled with the input impedance, comes out to 10kΩ.

To find the value of the resistor, first look

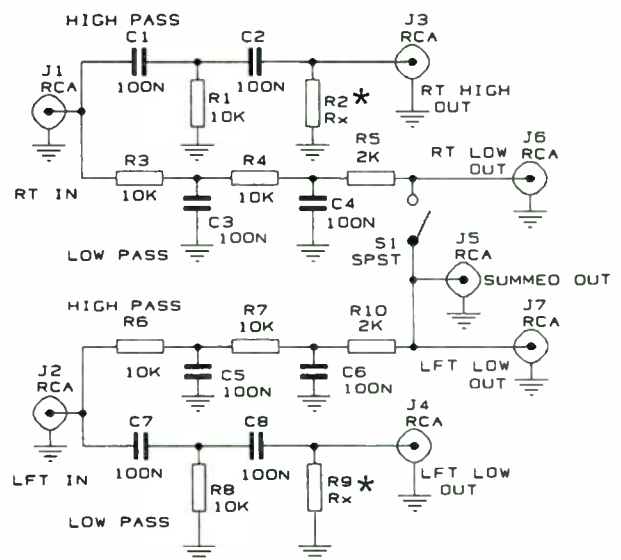


FIGURE 1: Passive line-level crossover for the Sipline satellite/woofer system (to be used between preamp output and amp input). Resistors: ¼W; Capacitors: monolithic or PC mount Mylar. *See text.

up the value of your amplifier's input impedance in your owner's manual. Enter that value in your calculator and press the 1/x key. Then press the minus key, enter 10,000, and again press the 1/x key. Now press the = key and the 1/x again. This will give you a minus number. If you ignore the minus sign, you will have the correct value of resistor Rx in ohms.

Example: amplifier input impedance is 30kΩ. 30,000 (1/x) 0.0000333 (-) 10,000 (1x) 0.0001 (=) -0.0000667 (1/x) -15,000. Ignore the minus sign and the answer is 15,000 ohms (15kΩ).

In spite of the fact that the crossover is second-order, and theoretically should function best when the satellite leads are reversed in relation to the woofer leads, I now find the system sounds better when the satellites are hooked up with the same polarity (plus to plus, minus to minus). This is probably due to the effect of the sound's different path lengths from the Simplines and the woofer to my listening position.

I suggest hooking up your system both

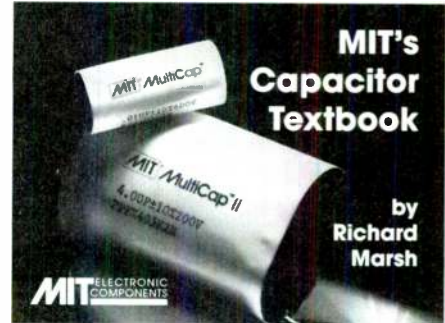
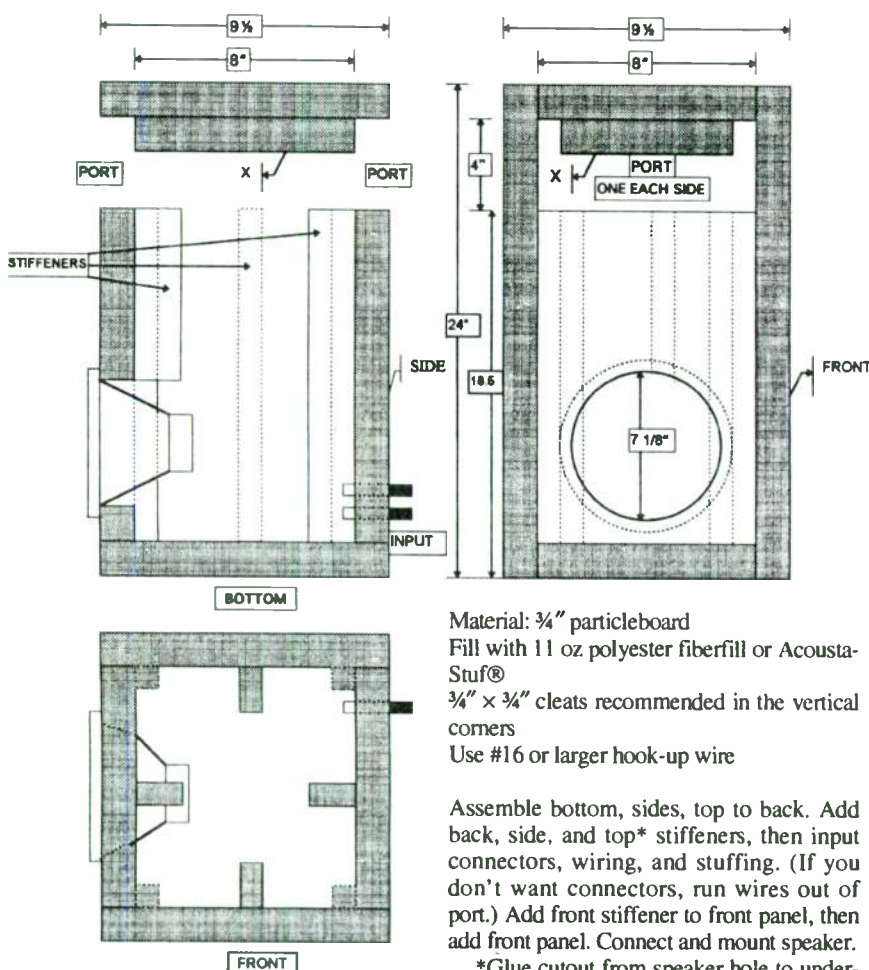
ways and retaining whichever polarity sounds best with your setup. Since most of the sound from this system is reflective, it is difficult to predict path lengths simply from looking at the setup arrangement. It might also be that, since I am using amplifiers from different manufacturers, one amplifier has its polarity reversed from the other. The key here is *whatever works!*

For use in a tape loop circuit, instead of between a preamp and an amp, a stereo volume control is required ahead of the crossover inputs because the record out jacks bypass the preamplifier volume control. The Radio Shack 271-1732 will work. (Don't enable the loudness control feature.)

I hope this gives you something to play with. Please write if you have any questions or comments. I'll probably keep this monstrosity hooked up. Thanks to Harry Campbell for that.

to page 56

CONSTRUCTION DETAILS



Debunking the Myths

Cap Myth #3

Some caps in audio applications sound fine without by-passing.

Cap Fact

All capacitors in audio applications need by-passing: coupling caps, crossover caps, power-supply electrolytics. Without by-passing, high-frequency performance and transient response will be compromised.

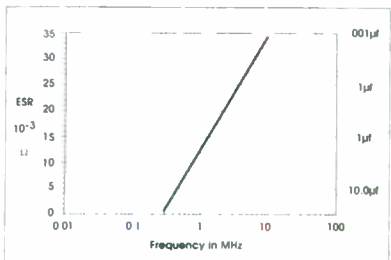
- **Problem:** Capacitors slow down at higher frequencies. The larger the capacitor, the slower the response.
- **Solution:** By-passing a larger slow cap with a smaller faster cap (MIT recommends 10% of the overall value) restores high-frequency performance.
- **Optimized Solution:** The patented MIT MultiCap - the only internally by-passed capacitor available. Each MultiCap is constructed by paralleling up to 10 sections, each 10% of the overall value.
- **Results:** Lower parasitics than possible with ordinary by-passing. Superior high-frequency performance in a single unit without additional, external bypassing.

The MultiCap is also an unequalled by-pass for power-supply electrolytics.

Watch for the next capacitor Myth!

"...the MIT capacitors... charge and discharge much faster. The speed of these caps also implies that they are passing only the signal present at that instant, leaving no residual effect from any previous signal."

From *The Audiophile Voice*, Vol. 2 Issue 1, p. 105 - a comparison of the ARC SP 9 Mk II & Mk III



ESR vs. Frequency vs. Capacitance

Graph showing low distortion obtained by the decreased parasitics (ESR) of the self-bypassed MultiCap. MultiCap performance falls to the right of the line, indicating fewer parasitics, i.e., fewer colorations & increased accuracy. Measurements of standard caps, even with exotic foils, falls to the left of the line, indicating compromised performance. (Measured at MIT)

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Speaker Builder 4/95 9

FOCUSED ARRAYS: MINIMIZING ROOM EFFECTS

By Bill Waslo

One of the most critical hardware components in an audio system is the listening room. Expensive electronics and loudspeaker components located in a poor room (or poorly placed in a good room) can, and often do, give less satisfying results than a less expensive system in an exceptional room.

Unfortunately, for many of us the room is often the least easily replaceable part of a stereo system. Our rooms are in our homes, which are part of neighborhoods, as are our kids' schools, and our jobs. We might think we have only a few choices in adjusting the room part of the system equation with room treatments (rugs, panels, absorbers), orientation of speakers and listening position, or by design of speakers and electronics to deal with particular room effects (floor-mounted woofers, in-wall speakers, equalizers).

A possible solution occurred to me when I read a recent AES journal article which

described a test in which listeners' preference was determined when speaker brand and position were varied. The room position was a more significant factor in perceived sound quality than the speaker model.¹

Interestingly, another article in the same journal described a totem-pole arranged microphone array for use in meeting rooms to achieve a directional pattern with main lobes in a donut shape peaking at the speaking positions around the room.² I thought if this works for microphones, it should work for speakers, too, and set out to discover an array of drivers to reduce the effect of room acoustics on loudspeaker sound quality.

MULTIPLE SPEAKERS

Driver arrays remind us of the marketing ploys often used to sell low-quality speakers ("Three midranges! Six tweeters! Four ports!"). Such designs prey on the "more is better" perception, and, if there is a technical

angle at all to the approach, it is usually toward improving dispersion. Some designs employ an array of drivers facing forward, backward, and/or sideways with the aim of increasing the amount of sound reverberating from room surfaces for a more diffuse sound and to increase the effect of the room's sonic signature.

My array is different; its purpose is to increase the intensity of the direct sound path while reducing that of the reflected paths, thus minimizing the sound characteristics contributed by the room. In a way, it is the antithesis of systems which enhance reflected energy, although both ideas use a driver array.

The use of multiple drivers to reduce reflections may seem counterintuitive at first, but radio antenna engineers have an analogous technique. They lower multipath distortion (which is a corruption of a direct radio signal by reflected signals) by utilizing directional antennas composed of an array of spaced elements. The antennas are advantageous in the transmit path, the receive path, or both wherein the array not only improves the echo rejection but also increases antenna gain or sensitivity to the radiated signal. The use of transducer arrays is also common in sonar and ultrasound technology.

OPERATION

So, how does it work? The concept is very simple, so I suspect it must have been thought of before. Yet neither I nor anyone I've talked to about this has seen or heard of such a system. Basically, each "speaker" consists of an array of drivers spread out on the front-left or front-right side of the room.

The critical characteristic is that each driver is placed the *same distance* from the listener's head (Fig. 1), but at different distances from the room boundaries and other reflecting surfaces. Ideally, they should be irregularly spaced from room surfaces to randomize the arrival times of the reflections at the listening position.

The direct signals arrive coherently (phase aligned) at the listening position, and thus the

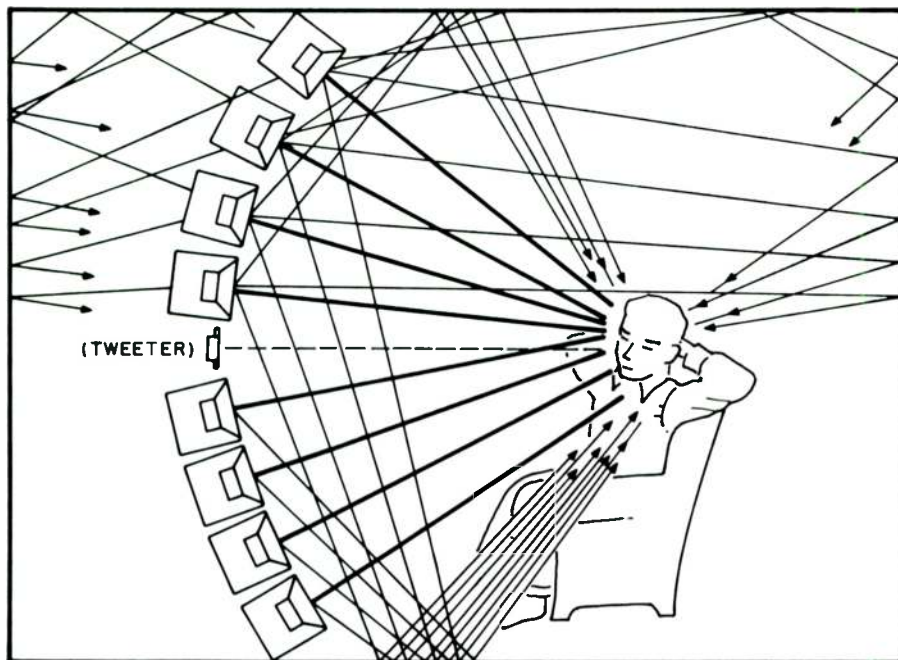


FIGURE 1: Direct signal paths are equal; reflected signal paths vary greatly. Only a fraction of the reflections are shown.

pressure sums directly (increasing 6dB for each doubling of the number of equally powered source drivers). The reflected signals, however, arrive at different times. Thus their strength sums (ideally) at 3dB for each doubling of equally powered sources. Two drivers arranged this way should have up to a 3dB (6 - 3) enhancement of direct over reflected intensity.

In general, the maximum decibel enhancement will go as $10 \log(N)$, where N is the number of drivers. Four phase-matched drivers should show 6dB of enhancement, and eight drivers can yield 9dB. This process is very similar to the noise-reduction-by-averaging scheme used in the basic IMP measurement system. However, reflection intensity, rather than noise reduction, is reduced relative to direct signal intensity (Fig. 2).

You can imagine that the listening position will be very critical. This setup is intended to be ultradirectional and foregoes the usually desired goal of loudspeaker omnidirectionality. But after spending thousands of dollars on equipment, doesn't an audiophile always sit in the one best seat, anyway? But you would not invite a large audio club over to listen to your focused array system. You can avoid the need to install a vise in your listening chair to precisely hold your head, if the higher frequencies (and perhaps some of the midrange) are not part of the array.

You could use a single tweeter of narrow dispersion characteristic in a traditional placement, or a controlled directivity (horn) tweeter, with other drivers arrayed widely around it. A single tweeter would also help preserve imaging and localization. You might notice the single tweeter, dual midwoofer setup as one of the characteristics of the D'Appolito configuration, which might be considered a "3dB enhancement" focused array for midrange fre-

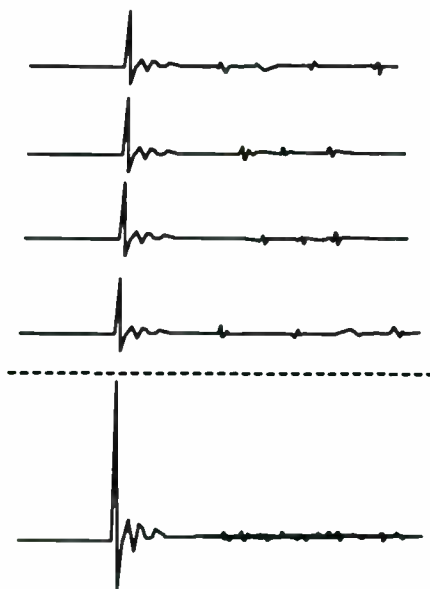


FIGURE 2: Impulse responses due to distributed drivers have the direct part aligned, but the reflected parts spread (top). The summation of these at the listening position reinforces the direct signal relative to the reflected signals (bottom).

quencies (although the D'Appolito configuration has other design goals).

IMPLEMENTATIONS

You can place the drivers in the arrays various different ways. You could install each into its own small cabinet or on its own stand, equidistant from the target listening point. This would locate them pseudo-randomly on the surface of an imaginary sphere centered at the listener's head as in Fig. 3. You could also locate some drivers at the sides or even behind the listener, but I suspect toward the

front would be more satisfying. Writing a computer program to help position the drivers could optimize the results.

You could combine the array into a set of less visually overwhelming "speakers" as a commercial product by installing them into long, inclined arc-shaped cabinets (Fig. 4). Other possible shapes of unified-box systems include X-shaped, S-shaped, and, of course, a full satellite-dish arrangement using many drivers. You might array your system with multiple midranges and multiple woofers for three-way, four-way, or more versions.

Some other variations you might consider are: arraying the drivers into a planar grid, with the arrival times aligned via electronic delays, or behind a porous screen of a projection video system. Another variation is to use ribbon drivers or electrostatic panels for the driver.

THE SOUND

If you are wondering what this will sound like, so am I. These days I spend more time with a computer keyboard than with a table saw, so I haven't tried this idea yet. I am presenting this approach to the *Speaker Builder* community to see whether I can tempt some of you. Here's your chance to possibly make modest audio history. Fire up your saws!

For a simple, first-time experiment, you could make about 16 satellite-sized mid-woofer cabinets, spread them out (in left and right sets), and arrange them for equal arrival times at ear level in your listening chair. The best way to align the drivers is by measurement of the impulse response as received at the listening position. A marked length of string is also usable for equidistant alignment. If you use a single tweeter for better sound, you may want to use separate amplifiers for midwoofer and tweeter for the test to more

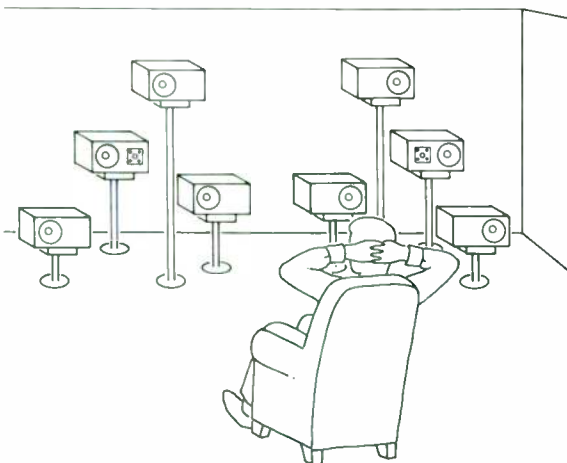


FIGURE 3: A "6dB enhancement" focused array using separate cabinets and stands.

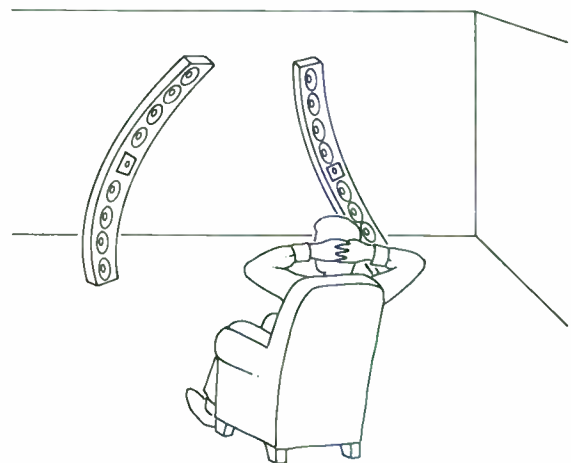


FIGURE 4: Inclined arc implementation of a focused array.

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easily match the levels of the drivers as you add or subtract midwoofers to the array. In the final configuration, a suitable series-parallel arrangement may be possible for the midwoofers to convert to a passive crossover.

SOME NAYS

There may be some negatives with the array concept. As G.R. Koonce points out, your current recordings may not necessarily sound best in a reflection-reduced environment.³ This is the chicken-and-egg problem with recordings and speakers. Recordings are artistically made creations, not simply objective measurements of a sound field. A recording engineer uses techniques and mixes which will sound best on the speakers that he anticipates will be interpreting the recording. Typical mixes probably assume room effects. Since I've never heard of the existence of a focused array system, I doubt that any recordings have been monitored or produced with them in mind.

Koonce also notes that reflected energy may affect the imaging illusion. Remove it and things could become kind of dry. You might wonder if the result would be much different than headphone sound. Of course, the reflection reduction will not be total, and it may be desirable to lessen the room's colorations, but not eliminate all reverberent fields, or to restore the field by separate rear or side channel speakers, home-theater style.

There exists one best seat, at one best distance from the system. The result is intended to be very directional, and not just the way a planar speaker is directional. On the other hand, I've often felt that the planar speakers' semi-coherent image, which often seems to come from behind me (reflected off the back wall), is quite detrimental to realism. The arrays might cure that problem yet retain the advantages of directionality.

Another problem might be the frequency dependence of the effect. Expect it to diminish at very low frequencies. Additionally, frequency-dependent sidelobes in the radiated pattern may lend a coloration to the remaining reflected energy.

Focused array systems could also become rather expensive. Eight decent midwoofers per side can add up quickly. Driver manufacturers would be delighted if this became popular, and so would speaker-stand manufacturers if audiophiles embraced the arrangement of Fig. 3.

AND YEAS

Since I'm writing the article, I get to finish by listing some of the advantages of focused arrays. In addition to a smoother response, expect the following desirable effects:



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A FLEXIBLE FOUR-WAY SYSTEM

By Ron McGhie

In recent years, recorded material has placed an ever-increasing demand on domestic (home) speaker systems with respect to low-frequency output at high levels. My system, operated at its rated input power, can exceed 115dB SPL over a relatively long period of time, while remaining docile enough for chamber music at reduced levels (*Photo 1*).

DESIGN GOALS

For as long as I can remember, I've been interested in loudspeaker design. Speaker building, if done properly, is one of the most satisfying projects you can undertake, rewarding the builder with sound that rivals the best commercial units. In some cases the results are better than commercial designs: when you build for yourself, you're not concerned with cutting corners to lower production costs.

If you use quality components, pay strict attention to detail, and, most importantly, be realistic about your speaker's performance, you can be proud of your efforts.

Most systems fall short of reproducing the dynamic range present in live performances...until now. This system addresses that shortcoming. Considering the small size of my listening room, it is virtually idling 95% of the time.

SYSTEM REQUIREMENTS

My system requirements, in order of importance, were:

1. Wide bandwidth;
2. High efficiency;
3. Modest-sized enclosure;
4. High power-handling capacity.

The system adheres to the following principles:

1. You can extend bass response with relative ease as long as high-output requirements are not taxing.
2. High system efficiency and/or high-input capacity do not guarantee high output at low frequencies.
3. Diaphragm movement to achieve a given level of acoustic output is a function of the frequencies, the size of the diaphragm, and the

type of system. It is not related to system efficiency unless the loudspeaker is burned out electrically trying to achieve the desired diaphragm movement.

Table 1 lists the peak-to-peak excursion in inches required to achieve one acoustic watt at the frequency indicated for vented (and sealed) systems. This shows the decided advantage of vented systems over sealed systems in terms of driver excursion at low frequencies. It also shows the benefits of choosing a larger driver over a smaller one when you require high outputs at low frequencies. In its simplest form, acoustic watts = $\eta_0 \times$ power rating (thermal limit). Note also that X_{MAX} specifications are one-half the peak-to-peak value mentioned here, and $\eta_0 =$ Thiele/Small reference efficiency.

So, how do you handle inadequate excursion? Well, you can keep peak acoustic levels below one acoustic watt, design around a larger-diameter speaker, design around mul-

iple speakers, raise the low-bass limit, or use a system type which minimizes excursion requirements. The all-horn system is the obvious choice, if not for the sheer bulk of the low-frequency section, so I opted for vented technology, which provides a reasonable compromise between size, efficiency, and low-end response (f_3).

SELECTED COMPONENTS

My initial design considerations centered around a two-way vented low-frequency section and a horn-loaded high-frequency section; that is, basically a 15" vented woofer with a horn "top end" capable of 400Hz–20kHz. While the low-frequency section was solidifying on my computer program, based on the JAES papers by A.N. Thiele and Richard Small, I discovered a pair of Electro-Voice drivers, called the VMR (vented midrange), which are used in the S1803 professional keyboard system. The VMR (EV part #1803-0302) has the same 16 lb magnet structure as the woofer I have chosen, the DL15W, which is a relatively new driver from EV intended for the professional sound reinforcement market.

For the midbass I chose an EVM10M musical instrument driver, which also has a 16 lb magnet structure. The sensitivities of these three drivers are very closely matched, eliminating the need for bothersome padding networks. I then chose the EV ST350B wide-angle tweeter, with 120° dispersion that is ideal for off-axis listening. Also, choosing drivers from the same manufacturer virtually assures they have the same "sonic fingerprint," which is impor-



PHOTO 1: The completed speaker system.

ABOUT THE AUTHOR

Ron McGhie is a telephone worker in British Columbia and has been involved in the installation and maintenance of electromechanical and digital long-distance telephone switching equipment. As an amateur musician, he has played clarinet in the Air Force Band and currently plays violin with the New Westminster Symphony Orchestra. Residing in Port Coquitlam, BC, he is the father of three and in his 47th year at the University of Osmosis.

tant when you are designing a "system." Driver specs are in *Table 2*.

ENCLOSURE DESIGN

Traditionally, loudspeaker system design begins at the low-frequency section, which is usually the bulkiest and dictates the size and shape of the enclosure. I needed an f_3 of at least 40Hz (normal mode approximately equal to a B4 alignment). Since the midbass enclosure is integral with the main enclosure, I actually needed to design two boxes. I started figuring how large to make this box, which must be added to the main box volume (*Figs. 1 and 2*).

My f_3 for the midbass was to be around 100Hz. My calculations showed a new volume of 0.68 ft³, yielding an f_3 of 93Hz and an F_B of 82Hz. With a 3" diameter PVC vent of 7.065 in² and a vent length of 1.97", driver displacement volumes for the EVM10M and the VMR are 210 in³ each for a total of 420 in³ or 0.24 ft³. So I added this to the 0.68 ft³ net box volume for a 0.92 ft³ gross volume.

For the low-frequency section, 5.1 ft³ yields an f_3 of 38Hz (normal mode) and 28Hz B6 assisted (equalized step-down mode). I added 500 in³ or 0.28 ft³ for driver displacement of the DL15W and 0.92 ft³ for

TABLE 1				
PEAK-TO-PEAK EXCURSION				
Loudspeaker DIA. IN.	50Hz	40Hz	30Hz	20Hz
8	0.77 (2.3)	1.2 (3.6)	2.1 (6.3)	4.8 (14.4)
10	0.50 (1.5)	0.78 (2.3)	1.4 (4.2)	3.1 (9.3)
12	0.31 (0.90)	0.48 (1.4)	0.86 (2.6)	1.9 (5.7)
15	0.19 (0.60)	0.30 (0.90)	0.53 (1.6)	1.2 (3.6)
18	0.13 (0.40)	0.20 (0.60)	0.36 (1.1)	0.81 (2.4)

the midbass enclosure for a gross box volume of 6.3 ft³. Because the design assumes no box losses, I added 0.4 ft³ (over volume). This resulted in a 6.7 ft³ box with outside dimensions of 36 × 24 × 17.5 (HWD). Normal-mode tuning occurs with 2–4" vents totaling 25.12 in² for a resultant F_B of 35Hz and a vent length of 6.71".

You can reconfigure this system to an alignment approximately equal to a no. 15 B6 assisted step-down mode resulting in half-octave extension in f_3 (from an f_3 of 38 to 28Hz) by reducing the vent area by one-half (i.e., plugging one of the vents). The F_B falls to 26Hz as a result. Then, by adding 6dB of boost at 28Hz with an accessory equalizer, you restore flat response with a new system f_3 of 28Hz.

this extra preparation for another project. After assessing how much of each size I had, I made the bottom, back, and subenclosure out of the 7/8" material and the rest out of 1-1/8" material.

I rough-cut all panels slightly larger than the finished size, then squared, made true, and labeled (right top, left bottom, and so on) the pieces (*Fig. 3*). I then cut them to size and dadoed and rabbeted the panels where necessary and routed the vents and speaker openings. The midbass enclosure shares both the midrange driver (VMR) and the midbass driver (EVM10M). They don't interact with each other, though, since the VMR is a sealed-back unit.

Photo 2 shows interior details before you add the midbass enclosure to the box. Due to the angles involved I deemed it easier to cut and fit the midbass enclosure as I went along. The integral midbass enclosure plays a major role in stiffening the main box; its bottom also serves as a front-to-back stiffening brace. Because this runs diagonally, a broader range of panel and box resonances are suppressed. *Photo 3* shows the finished midbass enclosure within the main box.

Photo 4 and *Fig. 4* show baffle details with the tweeter opening recessed to provide voice coil alignment with VMR. Left and right baffles are mirror images of each other, and the midbass, midrange, and tweeter are clustered as close as physically possible.

Liberal glue and screw down all joints

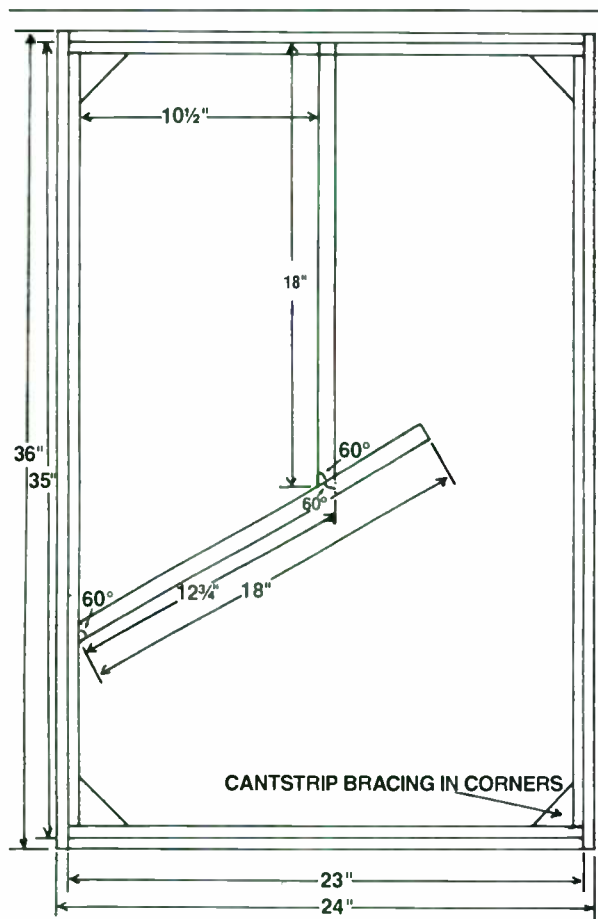


FIGURE 1: Front view with baffle removed.

CONSTRUCTION

The enclosure consists of 1-1/8" and 7/8" MDF particle-board. I acquired several sheets of what apparently used to be walls from a racquetball court. However, they were laminated on both sides with arborite, which I needed to skin off with a router wherever a glue joint occurred. Although these panels were extremely dense and heavy, I would probably not care to repeat

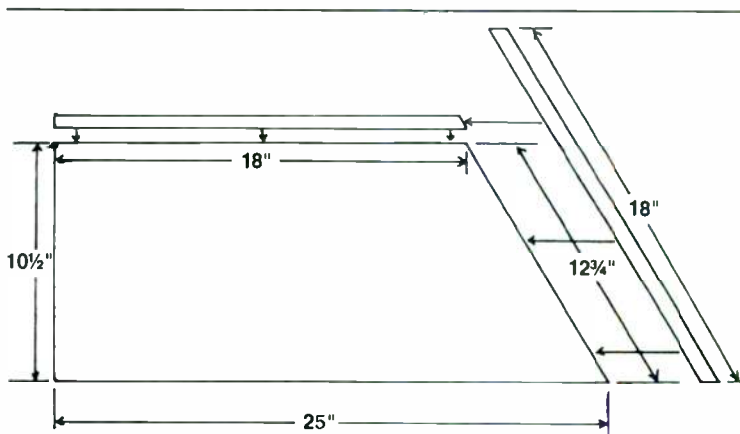


FIGURE 2: Midbass subenclosure.

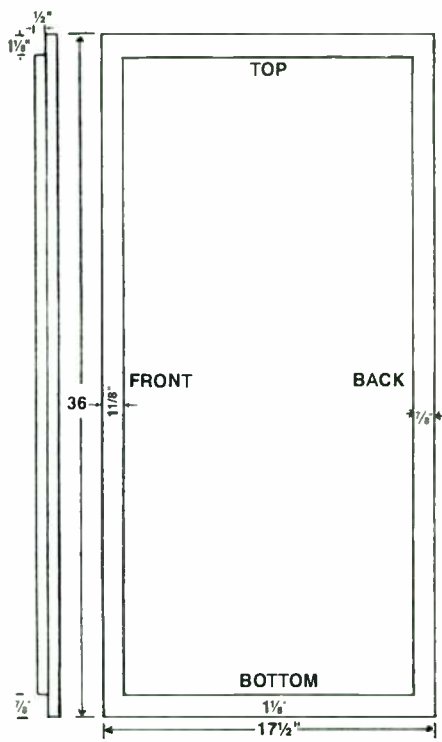


FIGURE 3: Sides.

along with the attendant panel bracing. Since the panels are not removable, access to the inside of the enclosure is through the speaker openings on the baffle. I caulked all interior joints with silicone sealant to ensure no box leakage. Cabinet damping is 2" fiberglass applied to three adjacent interior surfaces to make the boxes as inert as possible. As constructed, they weigh 220 lbs each with drivers installed. They also have minimal cabi-

net vibration when playing at loud levels and none when SPLs are more sane.

I used arborite to finish the top, bottom, front, and back, and will apply oak veneer to the sides later. The arborite gives a flawlessly smooth finish that is hard to duplicate by other means. I'll add a 1 1/2" base (not shown) before applying the veneer to the sides to prevent it from lifting when they are moved.

WIRING & TUNING

I designed the passive crossover and input panel with the idea that I would eventually biamp the system. The crossover points are 100, 350, and 3kHz. Of course, when biamping you can easily change the low-frequency crossover point. Figure 5 shows the input panel with straps and labeling for hooking up the system as an all-passive or biamp configuration.

Capacitors are Wonder-Caps® and inductors are all air core except for the 13MH and 24MH units, which are high-current ferrite core. Wiring is 10-gauge stranded throughout the crossover. I hot-glued the crossover components to the bottom of the enclosure and soldered them to terminal strips. The straps on the input panel make it

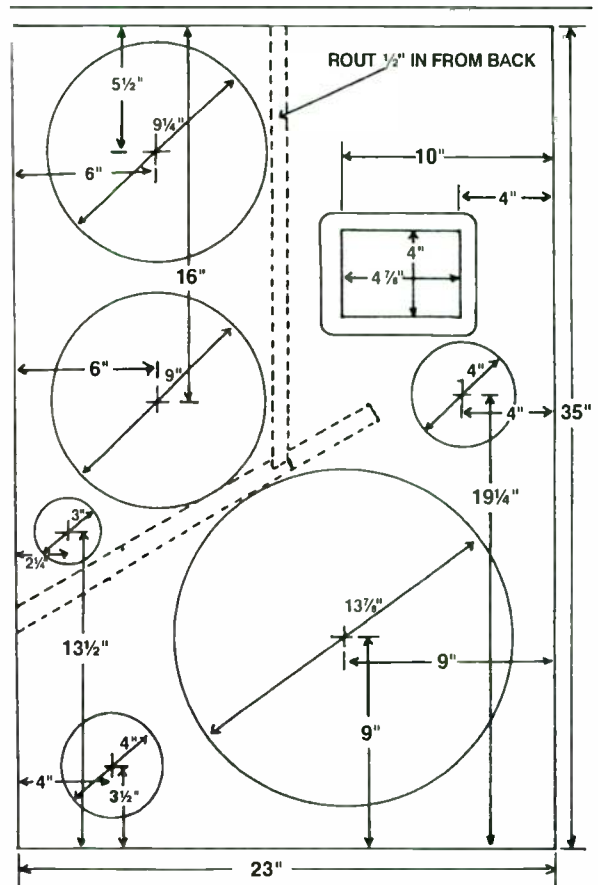


FIGURE 4: Baffle details.

easy to reconfigure from passive to active without opening the enclosure.

Wire individual drivers as noted in Table 3. You can easily reverse the DL15W polarity, since it appears directly on the input terminal strip. Of course, when biamping, you

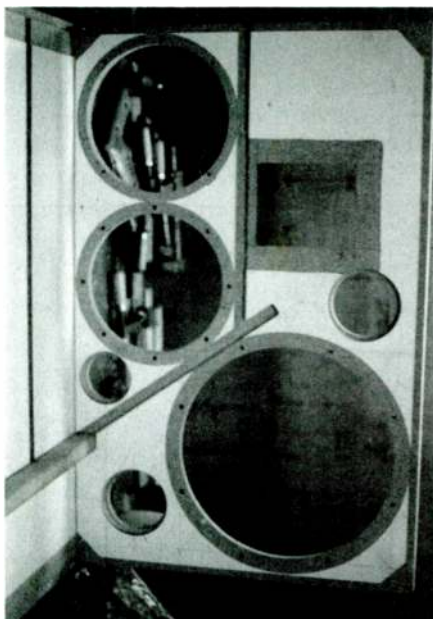


PHOTO 2: Inside view of box before mid-bass enclosure installation.



PHOTO 3: Midbass enclosure within main box.

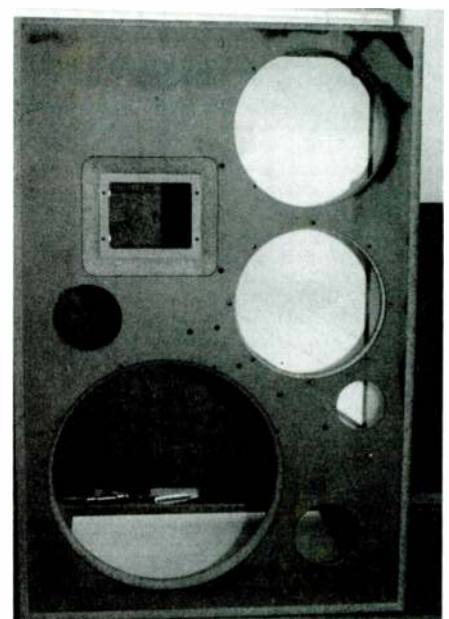


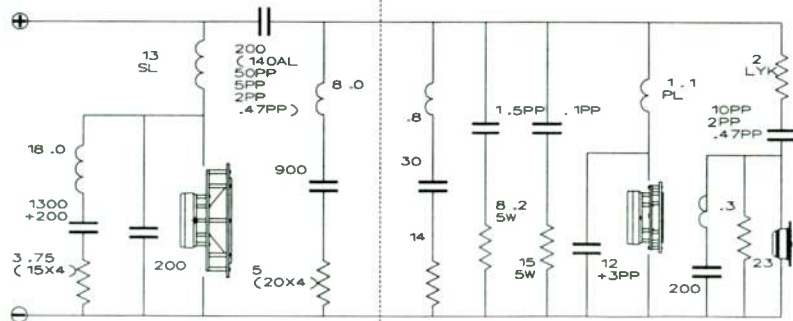
PHOTO 4: Front of enclosure showing driver and vent cutouts and recessed tweeter opening.

MIMESIS

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MIMESIS is for the discriminating listener who wants a system that has a flat frequency response, impedance magnitude that varies less than 2 ohms throughout the audio frequencies (less than 1/2 ohm for the satellite), extremely wide dynamic range, clean transients, and bass output below 25Hz. This is achieved in a sealed box design without equalization.

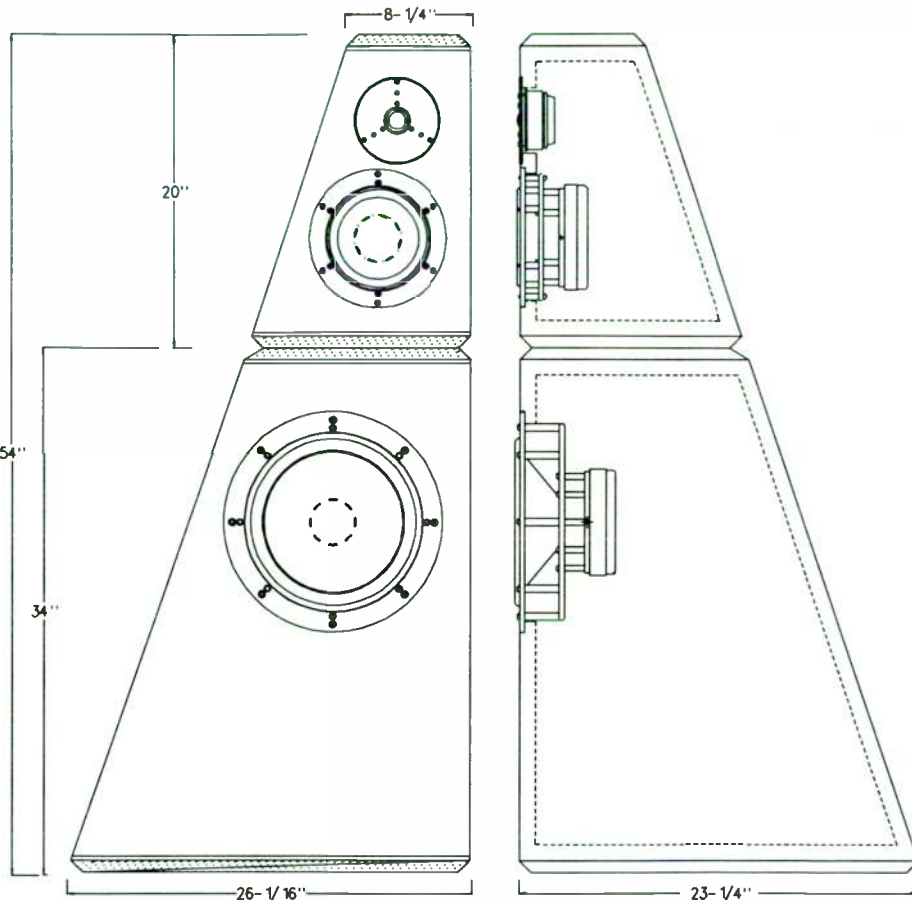
MIMESIS cabinet plans, drivers and speaker accessories are available from Madisound. For our cabinet, we choose 1-1/8" MDF with a cherry veneer and Corian tops and bottoms. Madisound stocks the SK300-304 12" woofer, SK170-308 6.5" woofer and the T330D Dynaudio Esotar tweeter. We can also provide you with Deflex acoustic panels for internal dampening, as well as input terminals and Flatline cable internal and external wiring.



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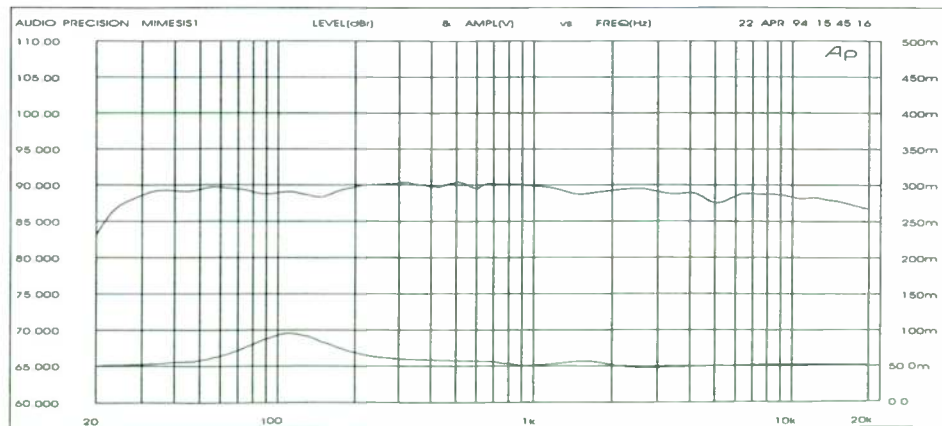


TABLE 2

DRIVERS AND SPECIFICATIONS

Low-Frequency:	Electro-Voice DL15W
Diameter	15"
F_S	30Hz
Q_{TS}	0.24
V_{AS}	8 ft ³
V_D	28.172 in ³
η_0	2.7%
Sensitivity	1W/1m/97dB
P_E	400W
X_{MAX}	0.22"
R_E	6.0Ω

Mid-bass: Electro-Voice EVM10M

Diameter	10"
F_S	65Hz
Q_{TS}	0.17
V_{AS}	1.4 ft ³
V_D	5.4 in ³
η_0	5%
Sensitivity	1W/1m/98.5dB
P_E	200W
X_{MAX}	0.11"
R_E	5.2Ω

Midrange: Electro-Voice VMR #1803-0302

With no published specs for this driver, I determined the following information:

Diameter	6.5"
Sensitivity	1W/1m/97dB
Power handling	50W RMS/500W PEAK
Dispersion	170° 500-1kHz

Horn tweeter: Electro-Voice ST350B

V/C diameter	1"
Frequency response	3k-15k ±3dB
Power handling	5W long-term average
Sensitivity	1W/1m/101dB
Horiz. disp.	120° @ 4kHz
	120° @ 8kHz
	110° @ 16kHz
Vert. disp.	90° @ 4kHz
	70° @ 8kHz
	50° @ 16kHz
R_E	6.2Ω
Directivity factor	$R_0(Q)$ 7.5 @ 4k

can achieve relative phase of the low-frequency section with a phase switch on the electronic crossover itself.

CROSSOVER COMPONENTS

Crossover components are available from many of the outlets familiar to *Speaker Builder* readers (Table 4). Don't substitute values unless, for example, you use two inductors in series to give you the desired value. You may choose to mount them in a fashion other than mine, but use the schematic as a wiring guide (Fig. 6).

Because of the difficulties of calculating a "true" net box volume (accurate measurements of driver displacements, crossover components, vent volume, and bracing can become involved, to say the least), a simple method for fine-tuning F_B is to make the vent longer (10% is usually enough) than

TABLE 3

RELATIVE PHASE CHART

ST350B	OUT
VMR	OUT
EVM10M	IN
DL15W	OUT

your design indicates. This noticeably lowers the box tuning below your target value.

By rocking an audio oscillator up and down below target F_B while holding a finger lightly on the woofer cone, you will find a null in the speaker's movement. This null occurs at box tuning. By applying the vent-end correction formula to the frequency where the null occurs, you will discover the proper length to produce your target F_B :

$$CHA L_V = CHA F_B 2L_V / F_B$$

where, $CHA L_V$ = required change in vent length in inches (negative value means reduce length, which will raise the F_B)

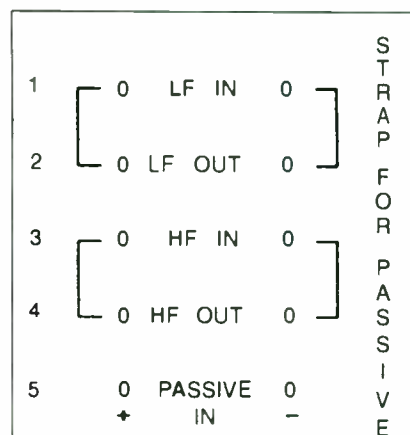
$$CHA F_B = F_B \text{ (required)} - F_B \text{ (actual)}$$

L_V = Initial vent length
 F_B = Initial box tuning frequency
 Depending on the accuracy of your oscillator, you can easily obtain results of <1dB error in F_B .

ACCESSORIES

This system accomplishes step-down equalization, when used, with the SEQ (EV's Sentry Equalizer unity gain device) in the Sentry III studio monitor, which provides 6dB of lift at a given frequency. Since the peak lift of the stock SEQ is at 28Hz, I did not need to modify it. By simply changing two resistors in each channel, you can change it to provide a peak lift at frequencies other than 28Hz. The SEQ, placed between the preamp and the electronic crossover, has the added benefit of acting like an infrasonic filter below cutoff. It has a three-position switch which provides high-frequency rolloff above 10kHz (0, -3, -6dB), depending on the "liveness" of your listening room.

My electronic crossover is an EV EX18 two-way stereo/three-way mono unit. The crossover points are continuously variable between 80Hz and 16kHz. It is a third-order (18dB/octave) device and level matching is



When operating in all-passive mode:
 Strap positive terminals 1 & 2
 Strap positive terminals 3 & 4
 Strap negative terminals 1 & 2
 Strap negative terminals 3 & 4
 Connect cables from amplifier to passive in terminals
 When operating in bi-amp mode:
 Ensure no terminals are strapped
 Connect low-frequency amplifier to LF in terminals
 Connect high-frequency amplifier to HF in terminals

FIGURE 5: Hookup instructions for passive configuration.

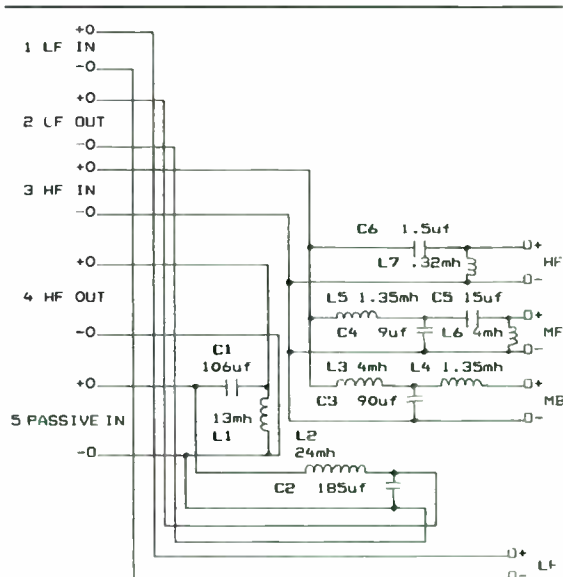


FIGURE 6: Passive crossover schematic.

TABLE 4

CROSSOVER PARTS LIST

VLF:	C1	106μF	100V min.
	C2	185μF	100V min.
	L1	13MH	High-current ferrite core
	L2	24MH	High-current ferrite core
LF:	C3	90μF	100V Solen Fast Cap or equivalent
	L3	4MH	Air core
	L4	1.35MH	Air core
MF:	C4	9μF	100V Solen Fast Cap or equivalent
	C5	15μF	100V Solen Fast Cap or equivalent
	L5	1.35MH	Air core
	L6	4MH	Air core
HF:	C6	1.5μF	100V Solen Fast Cap or equivalent
	L7	0.32MH	Air core

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SURROUND SPEAKER A 100

Bass reflex with slot-vent
H x W x D: 31 cm x 20 cm x 12 cm
(12.1/4" x 7.7/8" x 4.3/4")

Don't be fooled, this is a small but very potent home theater two way unit. You can use it as a front and/or rear speaker in any of the popular surround sound systems. The speaker mounted on a wall can generate a wide range of frequencies with surprising power thanks to a few special Focal Features:

- The 5N 412-DBL mid bass drive unit is one of the best Focal makes, combining dynamics (dual voice coil wiring) and natural sound.
- The TC 90 tweeter unit is of the latest generation of concave dome units with field replaceable moving assembly.
- The slot-vent open on the top of the enclosure couples with the wall and ceiling to naturally increase the bass output.



CENTER SPEAKER A 200

Bass reflex with front slot-vent
H x W x D: 18 cm x 50 cm x 22 cm (7.7/8" x 19.3/4" x 8.5/8")

The A200 center channel offers very low distortion, natural sound and high power handling. It features two 13 cm (5.1/4") midbass units of the same size and of the same cone technology as on the A 100, so that the same timbre will be produced harmoniously and consistently throughout your whole Home Theater system. The tweeter is also the same as on the A100 speaker, again to blend seamlessly with the other speakers. All units here are carefully shielded to protect your TV or your monitor from any magnetic interference coming from the very powerful magnets of the Focal units.



FOCAL

COMPACT B 300

Bass reflex with front slot-vent
H x W x D: 34 cm x 18 cm x 27 cm (13.3/8" x 7.1/8" x 10.5/8")
Certainly a small bookshelf speaker by its size, the B 300 is a real size performer, thanks to its unique dual voice coil mid bass unit and to the special crossover designed by Focal which uses the second voice coil in parallel with the first for a natural bass extension. In a 8.5l net internal volume (0.3 cu. ft.), this unique arrangement provides a -3dB response at 60 Hz! The tweeter is from the latest generation of Titanium dioxide concave dome tweeters with field replaceable moving assembly. A unique impedance adaptation device, positioned in front of the dome, controls its high frequency dome break up for an extended and sweeter, almost "silky", treble. This system will play loud enough for most mid size listening rooms, without ever become fatiguing. All crossover components have been carefully selected (we use high quality metalised polypropylene caps).



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PANEL CUTOUTS

Building a speaker system of this size requires a fair amount of wood. Therefore, careful layout procedures can minimize expense. Since not many of you will construct these speakers out of 1-1/8" and 7/8" material as I did, please adjust the dimensions in the construction plans to your material's thickness (Fig. A).

I suggest a minimum 1" thick MDF, but you may use 3/4" as long as you add panel bracing to the larger expanses to avoid resonances. You can't use too much bracing in an enclosure as long as you keep in mind that everything you add takes away from the final box volume.

You'll need just under 2 1/2 sheets of MDF for this project. You may use the leftover material for extra bracing or for another project. To make

each section more manageable, I suggest rough-cutting from A to A, B to B, C to C, and so forth. The dimensions are slightly larger than finished size. Once you cut all pieces as shown, trim them to finish size as shown on the enclosure plan drawings. Note that I chose to rout in 1/2" on those pieces which fit in this manner. If you choose not to do this, or to rout in more or less, adjust dimensions accordingly (Fig. B).

The finished sizes are:

Baffle: 23" x 35" (two required).

Back: 23" x 35" (two required). The back view is not shown since it is essentially the baffle without cutouts. Remember, how-

ever, to rout in the diagonal for the front-to-back brace, which is also the bottom of the midbass subenclosure. You also need an opening for the input recessed terminal strip made of leftover wood.

Bottom midbass subenclosure: 18" x 16.5" (two required). Cut the end that butts with the enclosure side at 60°.

Right side/left side: 36" x 17.5" (four required). If you rout the back of the midbass subenclosure into the sides, do so on only one right and one left panel as these are mirror-image devices. A simple butt joint will also work.

Back subenclosure: 11 1/4" wide (25" one

side, 18" one side) (two required). If you are not routing into the sides as noted above, adjust width by the depth of the rout.

Sides of subenclosure: 7 1/2" x 18" (two required). Angle end to butt against subenclosure bottom 60°.

Top: 17 1/2" x 23" (two required). Rout in as shown on opposite sides of these mirror-image panels.

Bottom: 17 1/2" x 23" (two required). I used cantstrip for the corner bracing, although other methods such as biscuit corners are acceptable. Refer to Bob Wayland's series on biscuit joinery (SB 4/94 and 5/94).

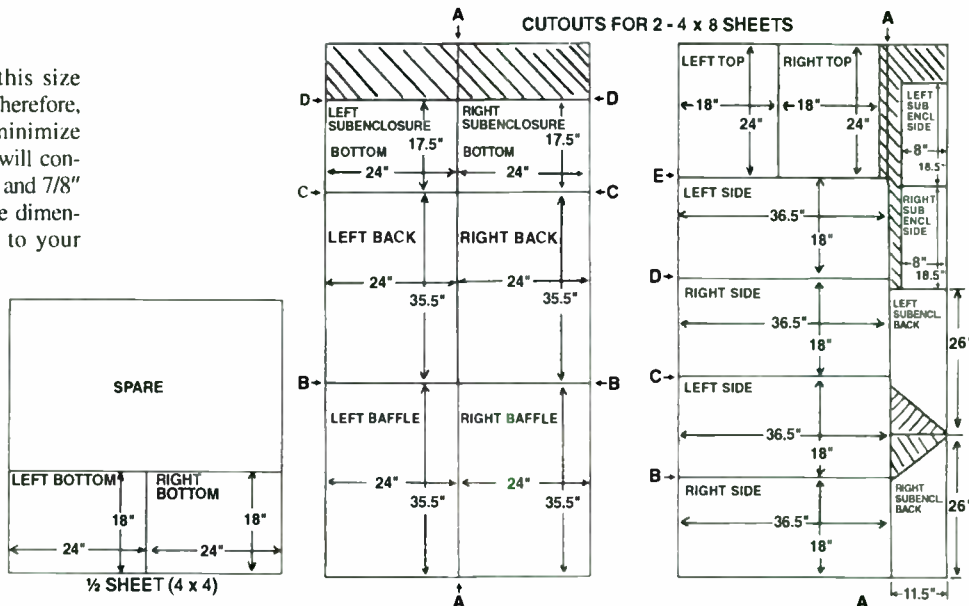


FIGURE A: Panel cutouts.

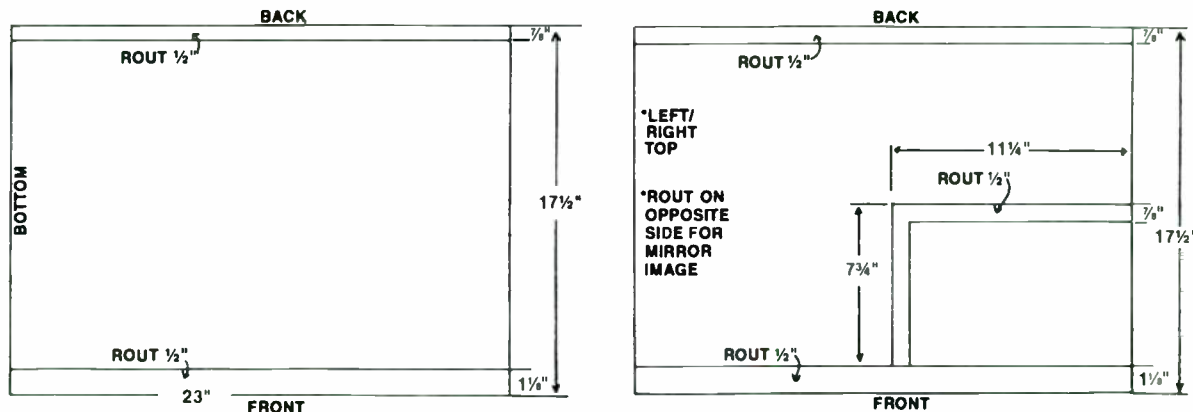


FIGURE B: Routing specs.

accomplished by increasing or decreasing the high-frequency output.

The addition of the active crossover, however, is another story. It will sound "different" than an all-passive system. I used a real-time analyzer (RTA) for initial level matching between the high- and low-frequency sections. Consider these as ballpark settings,

since you must use your ears for the final adjustments. By listening to varied program material, I found I was making minor level adjustments every week or so. Long-term listening is the only solution; your memory is capable of telling you what is too much and what is too little.

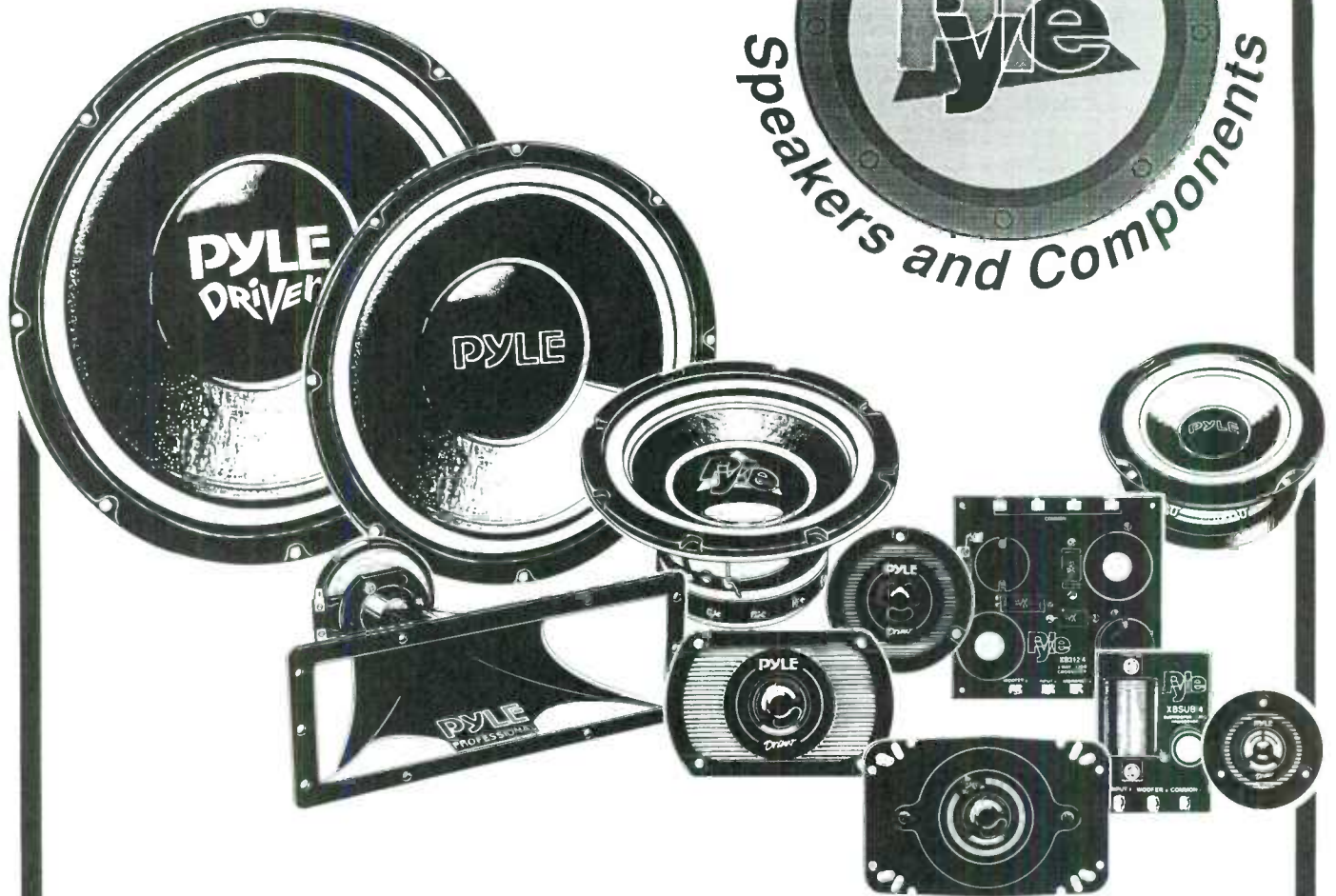
After a month or two, I used an RTA

again to see what the final reading looked like. Because of their size, I positioned the units away from room boundaries only for extended-length listening, then returned them to more "less imposing" locations for casual, everyday listening.

The more noticeable benefit of biamping, however, is the smaller load on the amplifiers

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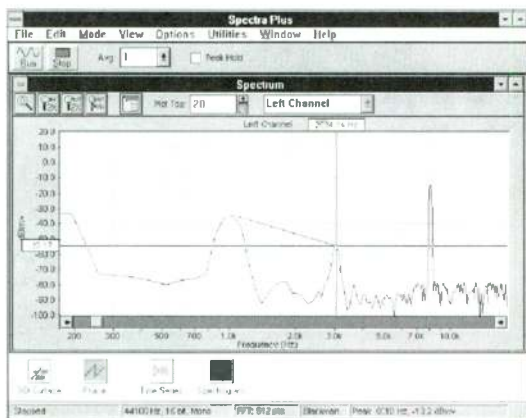
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when you band-limit them. IM distortion is not propagated into the driver hierarchy when you use a low-level crossover. This, along with an apparent increase in dynamic range, improved transient response, and improved amplifier/speaker coupling, results in a decidedly "cleaner" sound, free of problems with multi-way passive crossovers.

RESULTS

So, how did my project turn out? Well, the time I spent with the paperwork was worth the effort. I was pleased how well everything seems to work together, with no one section taking over. I suppose that is the way a system should work. I'm not ruling out the possibility of any tweaking down the road, but no glaring mismatches are evident so far.

The system's sound belies its appearance in that it doesn't sound as dramatic as you would think. But then that's what I set out to build. It enables you to listen "through" your music collection. Depending on the source material, however, the sound ranges from "quite up front" to "distant."

I find myself listening to the music rather than the speakers. When the power is applied, though, it can produce incredible SPLs. You really don't realize how loud it is until you try to talk to the person next to you.

Some might consider this overkill in a home system, but I believe I have more than adequately achieved what I set out to do, and if a system is not being pushed to its limits, it will sound better than one that is. In closing, my thanks to Ralph and Dave.

SUGGESTED READING

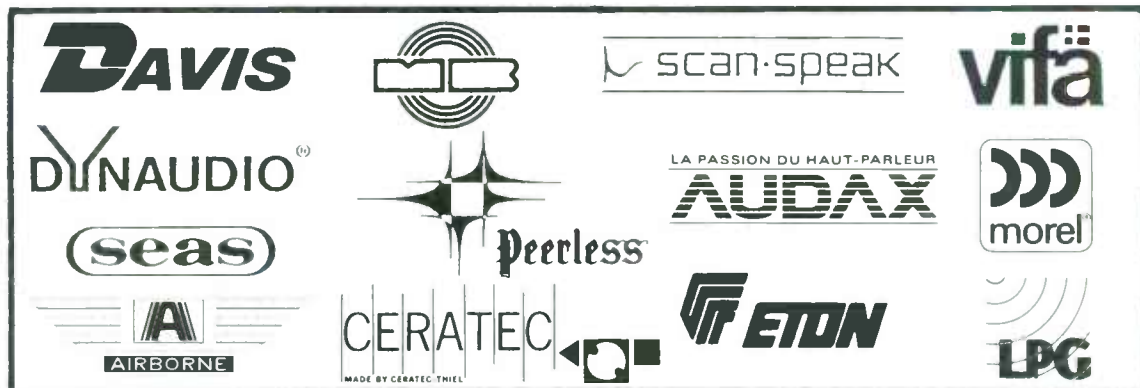
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2. R.H. Small, "Vented Box Loudspeaker Systems," JAES (Sept. 1973).
3. R.J. Newman and J.N. White, "Constructing Loudspeaker Systems With Predictable Performance," R-e/p (Aug. 1980).
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5. "Pro Sound Facts No. 7: DL Series Woofers—Plans For Recommended Enclosures and Small/Large-Signal Performance in These Enclosures," EV Publication (Oct. 1984).
6. "Pro Sound Facts No. 1: TL Bass Box Step-Down Equalization Through the Modification of a Sentry III Equalizer," EV Publication (Aug. 1975).

SOURCES

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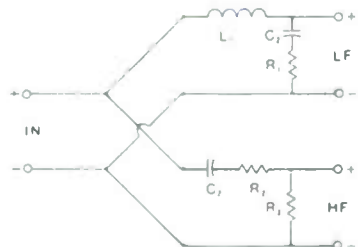


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EXTENDING IMP: A PROGRAM SET

By G.R. Koonce

The IMP Audio Analyzer by Liberty Instruments is a speaker testing device covered in several previous issues of *Speaker Builder*.¹ With its associated software it lets you run impedance and acoustic tests for drivers and speaker systems on your computer. You quickly learn that a

time-windowed tester does not solve all your problems with real-room testing. You must still carefully plan the test layout and use common sense to interpret your results. The IMP software also generates data files that other programs can process.

This article covers a set of programs that I

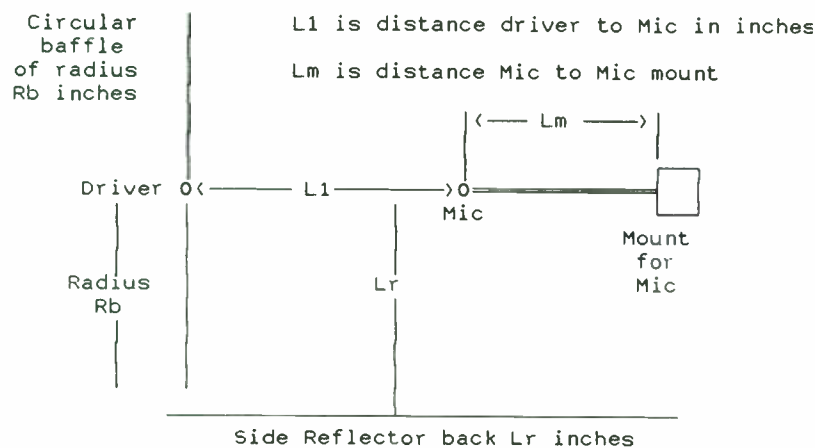
developed to work with IMP on an IBM-compatible computer. Two of the programs help you set up your tests, while the other three post-process data from IMP tests saved as files.

The Liberty Audiosuite software recently introduced by Liberty Instruments works with a sound card to provide testing capability exceeding that offered by IMP. Liberty Audiosuite will output files compatible with those generated by IMP that can be post-processed by the programs discussed in this article. While the article discusses IMP-generated files, these files could actually be generated by either IMP or Liberty Audiosuite.

ECHO-FREE TESTING

The purpose of time-windowed testing is to help eliminate room effects, which are normally reduced by near-field testing or test signals with some bandwidth to break up room resonances such as warble tone sinusoids or noise. Near-field testing, using fixed frequency sinusoids, simply overwhelms the room interference by a high signal-to-interference ratio. It works very well and is my favorite technique, but unfortunately, it has an upper frequency limit and

Model used by program:



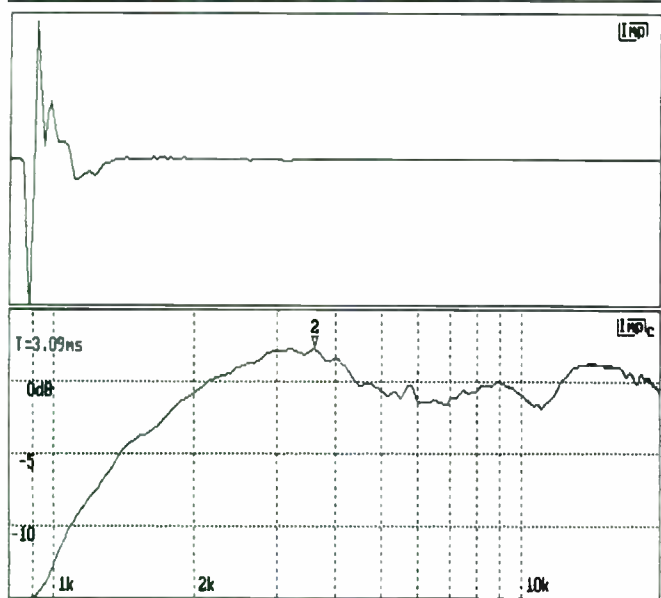
M to Menu, nul to next page:

FIGURE 1: Setup model used by IMPSpace. [In his programs the author follows the standard abbreviation for microphone. SB adopts an alternative style, mike.—Ed.]

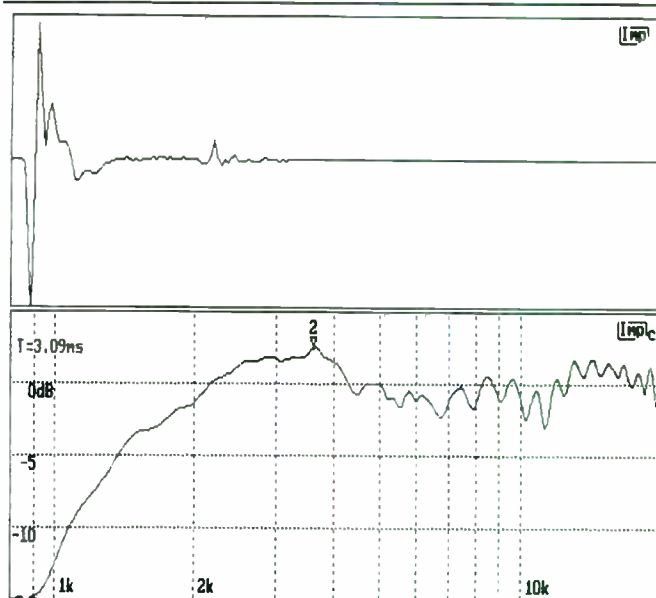
FFT window Tw =	2.20 mS	Min Freq of Interest Fmin =	1000.0 Hz	
Period of Fmin Tp =	1.00 mS	Ratio Tw/Tp =	2.2000	
Minimum Mic length to mount Lm =	21.70 inches		01-17-1995	
Dvr to Mic L1 inches	To Side Refl. Lr inches	Baffle Radius Rb inches	All echoes must be delayed mS	Delay FFT window mS
9	25.81	25.42	3.86	0.664
12	27.04	26.40	4.08	0.885
15 - - -	28.22	27.27	4.31	1.106
18	29.35	28.06	4.53	1.327
21	30.44	28.77	4.75	1.549
24 - - -	31.49	29.42	4.97	1.770
30	33.49	30.56	5.41	2.212
36	35.38	31.53	5.85	2.655
42 - - -	37.17	32.37	6.30	3.097
48	38.89	33.09	6.74	3.540
60	42.10	34.29	7.62	4.425
72 - - -	45.09	35.23	8.51	5.310
84	47.89	36.00	9.39	6.195
96	50.53	36.64	10.28	7.080

R to enter new data, nul to Menu:

FIGURE 2: Sample result from IMPSpace.



360.0Hz: -24.9dB -17.4° Δ -27.15dB 3600.Hz: 2.237dB 139.3°
FIGURE 3: Tweeter measurement with clean time response window.



360.0Hz: -23.4dB -20.5° Δ -25.95dB 3600.Hz: 2.475dB 141.6°
FIGURE 4: Tweeter measurement with echo introduced.

becomes very messy with multiple-output systems, such as multiple woofers, vented box, or passive radiator systems, covering the same frequency range.

The time-windowed technique avoids the room interference by processing the direct signal from the driver/system to the microphone and blocking room reflections by not processing data arriving after a certain time. This works well if you are careful with your test setup so that all echoes are sufficiently delayed.

Time-windowing's problem is that it puts a lower test-frequency limit on your result. You simply can't derive the 5Hz response of a system with only a 0.5ms sample of its output. Thus, you must make sure all objects that can generate an unwanted reflection are sufficiently far back, and use the proper window width with IMP for the desired results. The IMPSpace and IMPRoom programs help you set up your tests.

IMPSPACE PROGRAM

This program allows you to establish the requirements for an absolutely anechoic test of a driver: no unwanted echoes in the time window to be processed. You will probably only be able to establish such a test setup within a room with tweeters, i.e., drivers with the response down to about 1kHz. IMPSpace assumes you mount the driver flush in a baffle of adequate diameter to avoid the edge effects of the baffle, place all side reflectors back the necessary distance, and use a test microphone of sufficient length to avoid reflections from the microphone mount. *Figure 1* shows the model used by IMPSpace.

To use the program you enter the intended FFT window width (1–300ms), the minimum frequency of interest (51Hz–5kHz), and optionally the microphone-to-driver spacing (5–100"). If you do not enter microphone-to-driver spacing, the program will present a scan of spacings. The program indicates the minimum microphone length to its mounting, the minimum baffle radius, and the distance of all side reflectors from the test setup. IMPSpace will warn you if the width of the specified FFT window is incompatible with the desired test low-frequency limit. The program information section defines the rules the program uses.

Figure 2 lists the results for a test with a 2.2ms FFT window for 1kHz. Note that a microphone length of 21.7" is required to accomplish this test. For various microphone-to-driver spacings, *Fig. 2* includes the indicated time by which all echoes must be delayed, along with the approximate delay before the FFT window should start.

Figure 3 shows the time response (3.09ms) and the FFT-generated frequency response for a tweeter. This setup complies with the IMPSpace requirements, except no baffle was used, and the bare tweeter was suspended well off the floor, facing upward. The test microphone was 33" long from the end to the mount and hanging down from the ceiling. The time response is very clean, and so is the resultant unsmoothed frequency response.

Figure 4 shows the same tweeter in the same test setup, except a microphone stand is set back about 1' from the test path and introduces an echo at about 1ms. Note how this single echo source messes up the FFT-

generated frequency response of the tweeter by introducing numerous wiggles in the basic shape and making it difficult to identify true wiggles in the response.

The purpose of IMPSpace is to allow you to set up tests that produce clean results (*Fig. 3*). Note that you will not always obtain results this clean even with careful test methodology. I tried six different tweeter types in the same test setup, and only two produced time responses as clean as *Fig. 3*. The fault is not the test setup, but that many tweeters internally produce echoes, which are correctly considered part of the tweeters' response. If you are sure your test setup is clean, then you can identify such tweeters.

Remember, IMPSpace is designed to keep all reflections out of the test FFT window and is useful only with tests down to about 1kHz. As the FFT window becomes wider, you simply can't make the microphone long enough, or the baffle big enough, or position all the side reflectors back far enough. You must keep the microphone-to-driver spacing above a minimum for a valid response from the entire tweeter structure, which is developed in the next program.

IMPROOM PROGRAM

This program is based on the excellent article by Stuck and Temme,² who discuss the requirements to perform valid near-field and far-field tests. In general they show that you are best off using near field at low frequency and far field at high frequency. If all is well, you can overlap the two results to produce the total system response curve, which the IMP software is capable of doing.

IMPRoom offers the following:

1. For individual drivers, it will help establish the proper near-field test setup and test upper-frequency limit, both theoretical and real world.

2. For individual drivers or complete systems, it will help establish the proper far-field test setup and lower test-frequency limit.

3. For valid far-field tests, it will aid you in converting the result to a 1m reference, as is normally used.

When working with near-field testing of an individual driver, the program accepts a driver diameter input—either piston or nominal—and assumes the piston diameter is 80% of nominal. Optionally, you can enter the size of the structure (baffle or enclosure) housing the driver, which will affect the real-world near-field test frequency limit.

The program reports the maximum microphone-to-driver spacing for error less than 1dB, the theoretical upper test-frequency limit for 1dB error based on driver piston diameter, and (if you entered the structure size) the real-world test upper-frequency limit for 1dB error based on structure size. The decibel values to correct the measured response to a 1m reference are provided for 2π SR (steradian) and 4π SR radiation. Figure 5 shows a sample result for near-field testing, which does not require time-windowed measurement, so this portion of the program is useful for steady-state sine-wave testing.

For far-field testing of a driver, you again enter driver diameter, the dimensions for an entire enclosure, and the room dimensions. If your room supports an optimum valid far-field test of the driver/enclosure, then the program will display the test setup and limits (Fig. 6). The program includes the time delay before starting the FFT window, the window width, and the minimum valid test frequency for the indicated test conditions.

Here I disagree with Stuck and Temme, who use the minimum frequency as the period of the FFT window, which I believe is optimistic with IMP's rectangular window. So my program reports a frequency twice this value. You can determine your own value. Note that the output also provides corrections for converting the response to a 1m reference for a point source (decays 6dB per 2:1 distance increase free field) and a line source (decays 3dB per 2:1 distance increase free field).

If your room is not big enough to support a valid far-field test for your driver/enclosure with optimum echo suppression, then the program will offer the following options: you can either proceed with the best setup for the room—even if not a valid far-field test distance, in which case you can't convert to a 1m reference—or enter a microphone-

No Title Entered

Near Field Test Data & Limitations:

Driver Cone diameter = 6.800 inches.

Baffle dimensions: Height = 15.00 inches.
Width = 12.00 inches Thickness = 0.500 inches.

Maximum Mic to driver distance for 1 dB error = 0.374 inches.

Maximum theoretical test frequency for 1 dB error = 634.0 Hz.

Maximum real world test frequency for 1 dB error = 224.3 Hz.

Note: Use an FFT window width at least 2 times the period of the minimum frequency of interest.

To convert Near Field SPL to 1 meter reference:
For 2π sr free field radiation Subtract: +27.29 dB
For 4π sr free field radiation Subtract: +33.31 dB

FIGURE 5: Sample output from IMPRoom for near-field test.

Demo Plot of Testing Enclosure in Large Room

Far Field Test Layout & Limitations:

Room Dimensions: Length = 20.00 feet
Width = 18.00 feet Height = 12.00 feet

Test setup for Enclosure along center line of room's length:

Front of Enclosure out 7.071 feet from one end wall.

Mic out same distance from other end wall.

This gives Mic to source distance of 5.858 feet.

Center of front panel is set at 6.000 feet off the floor.

Enclosure dimensions: Height = 1.500 feet Width = 1.000 feet
Depth = 0.750 feet.

FFT window should start at about 5.190 ms.

Maximum FFT window length is about 6.641 ms.

Minimum valid frequency for test is about 301.1 Hz.

To correct result to a 1 meter reference:
For Point Source correction is +5.03 dB
For Line Source correction is +2.52 dB

FIGURE 6: Sample output from IMPRoom for far-field test.

to-driver/enclosure spacing to give valid far-field results, which may severely limit the low-frequency test limit due to early echoes. You can play around with various spacings until you achieve acceptable results.

IMPRoom has one further capability: when you enter the test distance used for a far-field measurement, the program calculates the decibel conversions for a 1m reference, as described previously. The program also allows you to enter data in metric or US standard units. All output screens let you display the

results in inches, feet, centimeters, or meters, and whichever dimension you choose will be used in printouts. This program has mouse support but can be keyboard-operated.

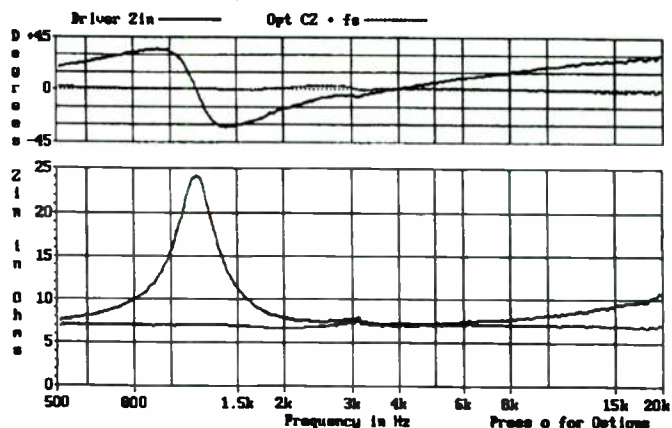


FIGURE 7: IMPZobel Z_{IN} plot for tweeter 1 with Complex Zobel with f_s correction.

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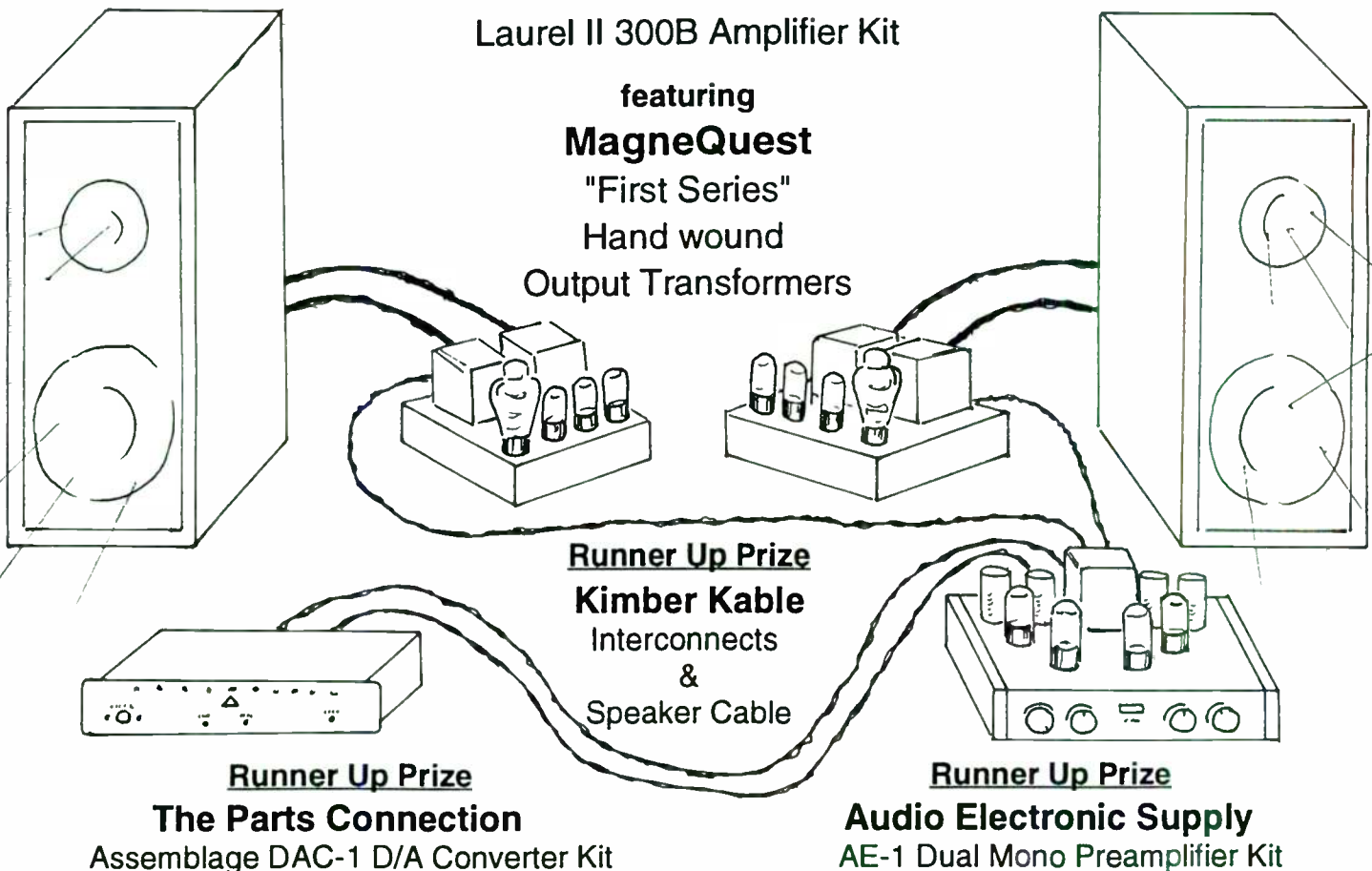
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Philadelphia Audio Society
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Devon, PA 19333

Colorado Audio Society
1941 S. Grant Street
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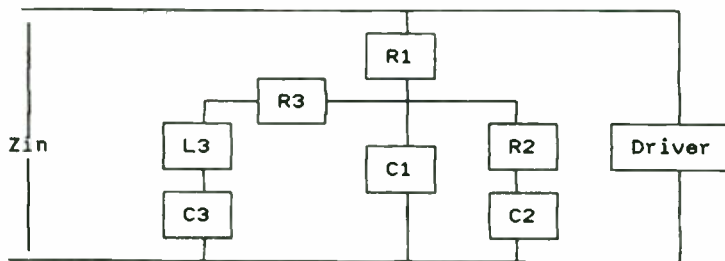
IMP is capable of outputting frequency response and impedance data in ASCII (text) files, with the .FRD and .ZMA filename extensions, respectively. They cannot be called back into IMP but can be used by other programs. If you use the original IMP software, you should save the time response files (including the Cal response) so IMP can retrieve the files and generate new .FRD or .ZMA files if needed. If you use IMP/M software or Liberty Audiosuite, you can save .FR2 or .ZF2 files, which can be recalled, but the following programs will only work with the ASCII .FRD and .ZMA files.

IMPZOBEL PROGRAM

This program works with an IMP-generated .ZMA file to develop Zobel for drivers. It will develop a range of Zobel from the simple, usual two-component variety to a complex seven-component variety that will flatten the driver impedance around the resonance range right up through high frequency. The complex Zobel are based on Victor Staggs' excellent article on driver impedance (*SB 5/94*, p. 28). IMPZobel works with IMP-generated .ZMA files but will allow keyboard entry of data sets (frequency in hertz, impedance magnitude in ohms, impedance phase in degrees), with a minimum of five sets

Complex Zobel for Dome Tweeter
Driver Zin Data from: TWEETER1.ZMA
Complex Zobel with f_s Correction:

Schematic for Complex Zobel with f_s Correction



Component Values: R1 = 7.000 ohm R2 = 4.296 ohm R3 = 2.865 ohm
C1 = 0.253 uFd C2 = 0.822 uFd C3 = 11.35 uFd
L3 = 1.647 mHy

Driver Data: Re = 7.000 ohm fs = 1,164 Hz Rmax = 24.10 ohm
Qms = 4.204 fm = 4,000 Hz

Model Data: Ra = 11.41 ohm Rb = 17.10 ohm Ca = 33.62 uFd
La = 0.040 mHy Lb = 0.012 mHy Lc = 0.556 mHy

FIGURE 8: Schematic for Complex Zobel with f_s correction for tweeter 1.

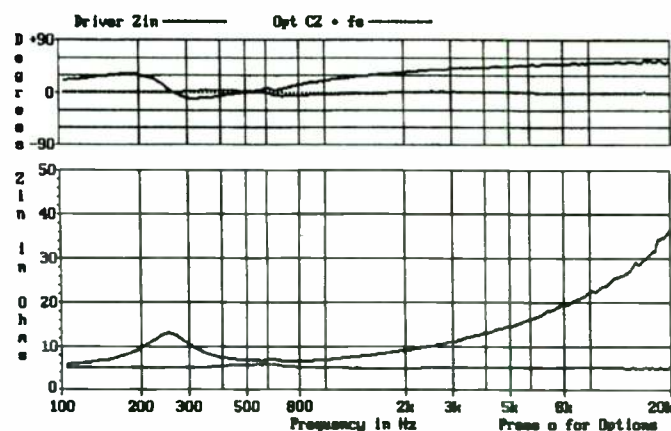


FIGURE 9: IMPZobel Z_{IN} plot for midrange with Complex Zobel with f_s correction.

required and a maximum of 50 sets allowed. You can write keyboard-entered data to a .ZMA type file for future use.

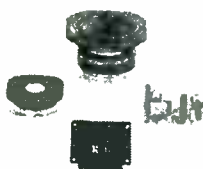
The program prompts you for the path to the .ZMA files, and then displays a directory of this path. You can select the proper .ZMA file by number. File size must be no greater than 30K; if the file is rejected as too big, call your original data back into IMP and generate a file which covers a narrower frequency span or reduce the file size as discussed later in this article.

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Value μF	Diameter mm	Length mm	SRP US\$	Value μF	Diameter mm	Length mm	SRP US\$
1.0	11	21	1.23	12.0	25	33	3.56
1.5	12	22	1.44	15.0	25	38	4.18
1.8	13	22	1.49	20.0	29	38	5.16
2.2	15	22	1.58	24.0	29	43	5.98
2.7	14	25	1.67	30.0	32	43	7.30
3.0	15	25	1.73	33.0	32	48	7.74
3.3	16	25	1.78	41.0	35	48	9.32
3.9	16	25	1.83	50.0	37	53	10.96
4.7	18	27	1.96	51.0	37	53	11.16
5.6	18	30	2.10	56.0	39	53	12.00
6.0	19	30	2.20	62.0	39	53	12.98
6.8	20	30	2.33	75.0	43	58	15.12
8.0	20	33	2.91	82.0	45	58	16.28
8.2	21	33	2.97	91.0	47	58	17.50
9.1	22	33	3.08	100.0	49	58	18.76
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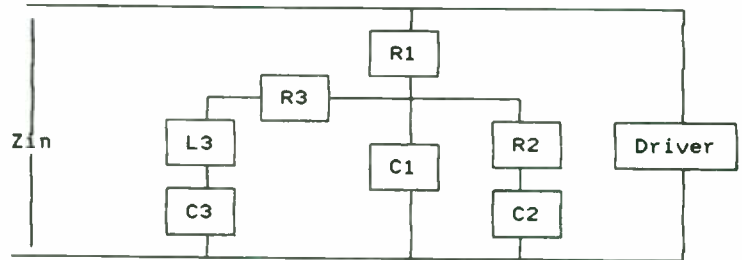
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Driver Zin Data from: MIDRANGE.ZMA

Complex Zobel with fs Correction:

Schematic for Complex Zobel with fs Correction



Component Values: R1 = 5.070 ohm R2 = 4.288 ohm R3 = 3.241 ohm
 C1 = 10.53 uFd C2 = 15.86 uFd C3 = 83.15 uFd
 L3 = 4.541 mHy

Driver Data: Re = 5.070 ohm fs = 259.0 Hz Rmax = 13.00 ohm
 Qms = 2.280 fm = 510.0 Hz

Model Data: Ra = 5.995 ohm Rb = 7.930 ohm Ca = 176.7 uFd
 La = 0.408 mHy Lb = 0.271 mHy Lc = 2.137 mHy

FIGURE 10: Schematic for Complex Zobel with f_s correction for midrange.

Once the file is accepted, you enter a frequency range to process/plot that can contain all or simply a portion of the frequency range of the IMP data file. You may then select a number of points to process up to the maximum in the file falling within the process/plot frequency range.

If you use too many points, the program will take a very long time to run. You might wish to start with only 20 or so points and then increase the number later for better plots. I find 150–200 points produce very good plots. The program has an information screen to aid you in picking the proper portion of the impedance curve for Zobel development. The program also allows entry of an optional driver name to appear on the plots and printouts.

IMPZobel provides four approaches:

1. Specified Simple Zobel: series R-C across the driver terminals, where you supply the component values. This allows evaluation of a Zobel design you already have. It is equivalent to simply trying the values and testing with IMP, except you don't need the various components.

2. Optimum Simple Zobel: series R-C designed by the program for best performance. The program varies R and C to produce the lowest phase error squared for the portion of the driver Z_{IN} curve with positive phase shift.

3. Specified Complex Zobel: you enter the values for the four components (two resistors and two capacitors) constituting a Complex Zobel. Again, this is equivalent to simply trying the values with the driver and testing via IMP, but you don't need the components. Attempting to determine the proper Complex Zobel values by trial and error can be frustrating.

4. Optimum Complex Zobel: the program tries to model the driver using Victor Staggs' schematic. If the model is successfully generated, you can develop three different Complex Zobels:

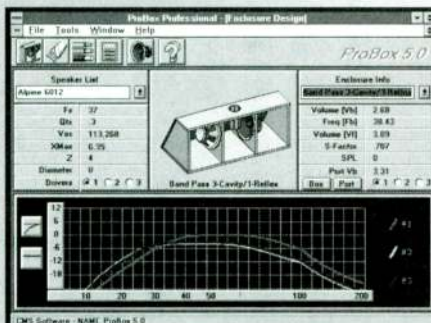
(a) normal Complex Zobel with two resistors and two capacitors to flatten the driver impedance at high frequency;

(b) network to correct driver impedance about resonance f_s . Network is a series R-L-C;

(c) combination of Complex Zobel with f_s correction to flatten the driver impedance over its full-frequency range. This network consists of the seven components composing the two networks above, but not all component values are the same.

The development of a Simple Zobel is common knowledge, so I will not discuss it

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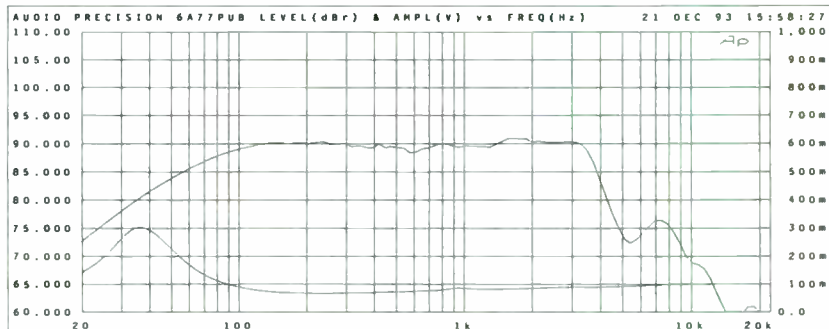
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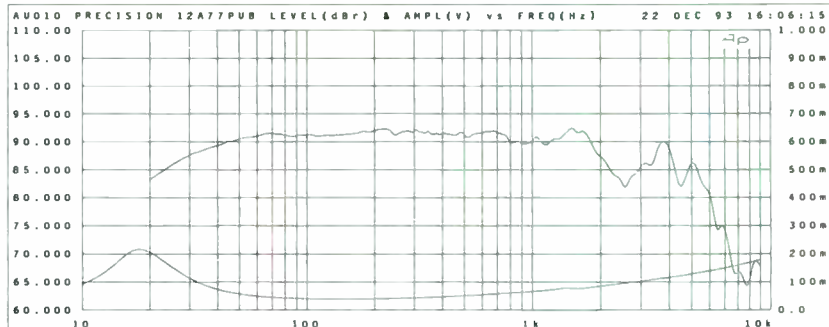
Technical Data	Symbol	SK170-308	SK300-304	Unit
Nominal Impedance	Z	8	4	Ω
Resonance Frequency	Fs	37	18.38	Hz
Power Handling Nominal	P	200	350	W
Sensitivity (1W/1m)	E	90	91	dB
Voice coil Diameter	\emptyset	77	77	mm
DC Resistance	Re	5.49	3.4	Ω
Voice Coil Inductance	Lbm	.243	.557	mH
Voice Coil Length	h	10	30	mm
Former		Aluminum		
Wire		Aluminum		
Number of Layers	n	2	2	-
Basket		210mm \emptyset Die Cast Aluminum	350mm \emptyset Die Cast Aluminum	
Cone Material		Mineral Filled Polypropylene		
Surround Material		Rubber		
Magnet Size		170mm OD x 20mm H, 77mm ID	170mm OD x 24mm H, 77mm ID	
Force Factor	BL	7.4737	9.5235	NA ⁻¹
Height of Magnet Gap	He	20	10	mm
Linear Excursion peak	Xmax	5	10	mm
Suspension Compliance	Cms	1161.44	1052.72	μmN^{-1}
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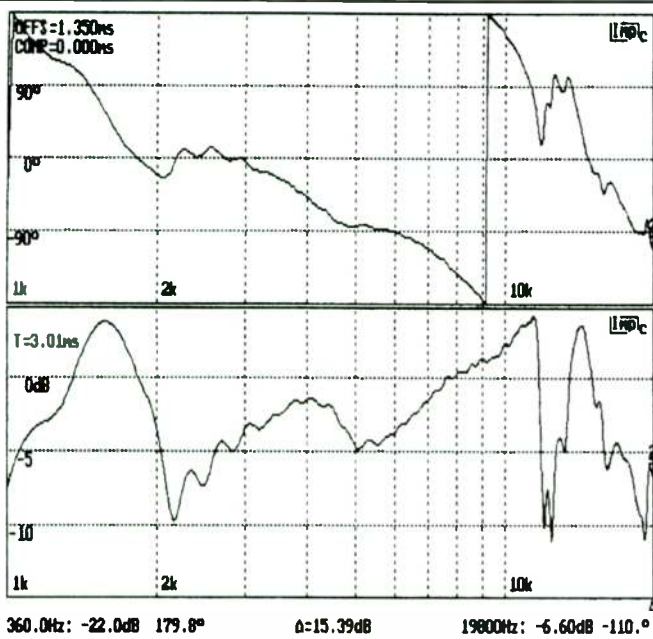


FIGURE 11: IMP frequency response plot of tweeter 2.

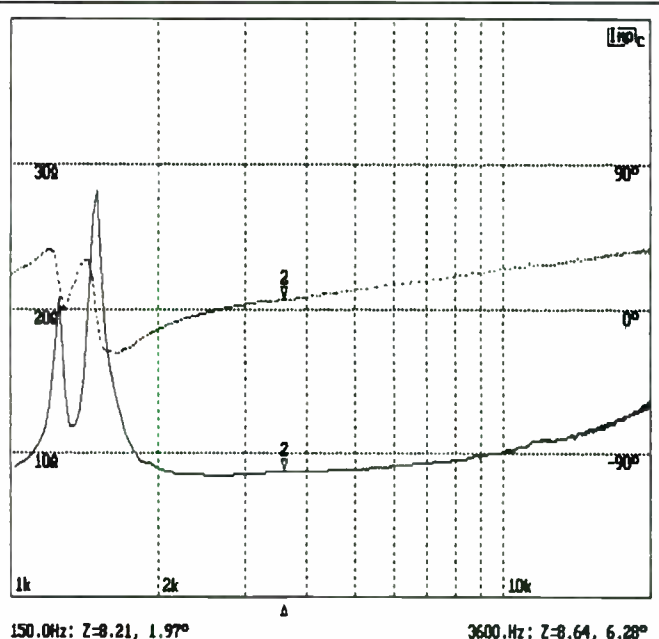


FIGURE 12: IMP impedance plot of tweeter 2.

further. The Complex Zobel is another matter. While the Simple Zobel is based on the driver assumed represented by a series L-R at high frequency, the Complex Zobel includes the effects of eddy currents and is far better at correcting the driver Z_{IN} to a pure resistance.

COMPLEX ZOBLER

With IMP, measure the driver impedance from below f_s to a high enough frequency that the impedance is rising. Then output a .ZMA file for this data. Using the markers for IMP, locate and record the resonant peak

resistance (R_{MAX}) and the frequency (f_M) where the phase goes through zero above resonance.

Now perform a T/S extraction to obtain the driver DC resistance (R_E), driver resonance (f_s), and driver mechanical Q (Q_{MS}). Print or record the T/S parameter data. With IMPZobel, import the .ZMA file and select Design Optimum Complex Zobel. The program prompts for the extra data listed above, then tries to model the driver Z_{IN} curve, and, if successful, allows design of the three networks described above.

It is possible the model generation

loop will fail to properly converge. The program will report this, and, if the error is not fatal, will allow you to proceed with using the model. You should plot the model Z_{IN} and driver Z_{IN} to see how well the model fits the driver before using any network designs. Certain loop failures are fatal, and you can't use the model, since network design equations involve dividing by zero.

You should try again with a different number of data points or a different frequency range portion of the driver Z_{IN} curve. The program offers information screens on what portion of the driver Z_{IN} curve to use for

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- Excellent Power Handling

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 Power Handling 60 W
 Qts 0.36
 Vas 29.26 L
 Voice Coil 35.5 mm
 Magnet 18 Oz.



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 Power Handling 60 W
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 Voice Coil 25.5 mm
 Magnet 5 & 10 Oz.



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DOB165R-PP/8SC 6-1/2"

- Double Magnet with Cup
- Home Theater Center Channel
- Computer Multimedia
- Vented Enclosure

Specifications

Impedance 8 Ohm
 Resonance 36 Hz
 Upper Frequency 6,000 Hz
 SPL 1W/1M 86 dB
 Power Handling 80 W
 Qts 0.30
 Vas 35.28 L
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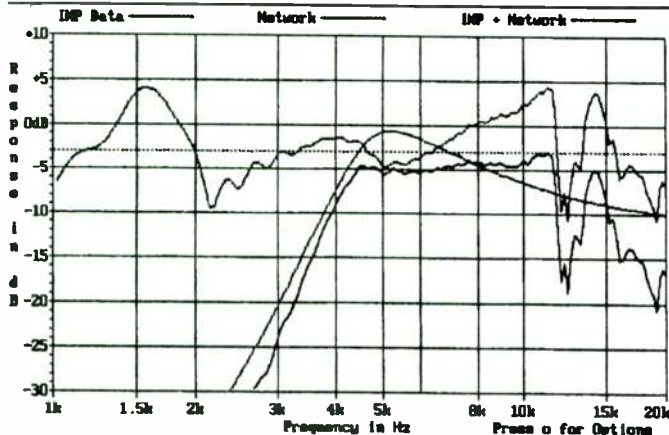


FIGURE 13: IMPPCoEq frequency response plot for tweeter 2 with equalization.

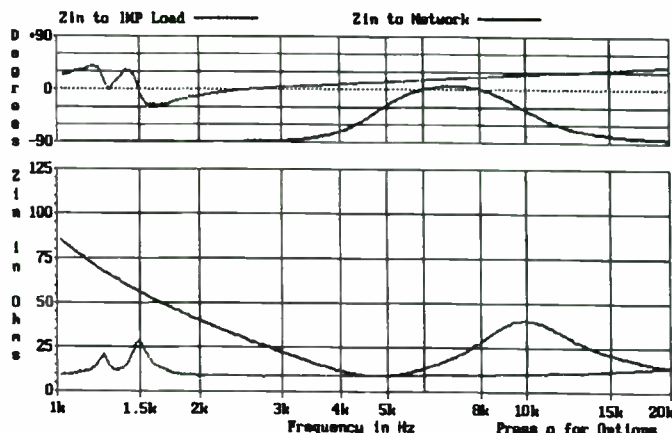


FIGURE 14: IMPPCoEq Z_{IN} plot for tweeter 2 with equalization.

development of Optimum Simple and Complex Zobel, and also shows all the equations used in the computations.

Figure 7 plots the raw driver Z_{IN} for a dome tweeter and Z_{IN} with a Complex Zobel with f_s correction; the result is an impressively flat and nearly resistive input impedance. Figure 8 shows the network schematic and component values along with the input data and model component values. The program will display and print the model schematic.

Figures 9 and 10 show the corresponding information for a drum midrange driver. Here it is clear the high-frequency impedance rise of the driver has been effectively canceled. Note on Fig. 9 that the IMP file contained 1,327 data points, but I used only 200 in the computations and plots. Again carefully consider the number of data points you use to avoid an unmanageably large impedance data printout. IMPZobel allows

you to plot the input .ZMA file alone, which, if it is the only data you have, might sometimes be useful, since IMP does not recall .ZMA files.

IMPPCOEQ PROGRAM

The IMP Passive CrossOver and Equalizer program takes the frequency response and input impedance data for a driver or system via .FRD and .ZMA files and places a passive network ahead of the driver/system for doing crossover (CO) or equalization. The passive network is composed of four branches: a serial branch (Z_a) at the input, then a shunt branch (Z_b) to ground, a second series branch (Z_c), and finally a shunt branch (Z_d), with the .ZMA file impedance across this last shunt branch.

The program plots the frequency response for the IMP .FRD data, the network from input at Z_a to output across Z_d (and across

the .ZMA impedance), and the combination of the network with the .FRD response. This shows what the output of the driver/system represented by the .FRD file would be with this passive network. The program plots input impedance (magnitude and phase) to the .ZMA file and into the network with the .ZMA load. It also prints tabulated data for frequency response or impedance (again, be careful the number of printed pages doesn't become too large).

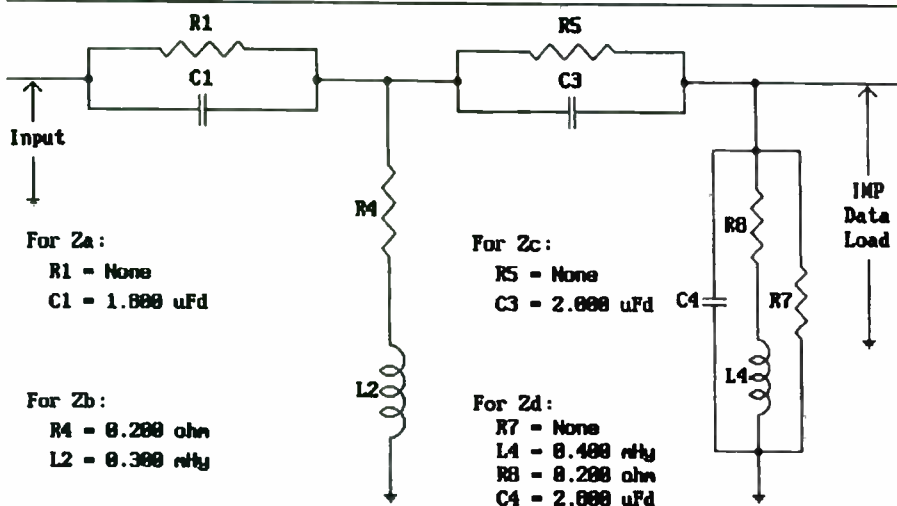
Each branch of the network may be:

1. a short circuit to omit the branch, if Z_a or Z_c ;
2. an open circuit to omit the branch, if Z_b or Z_d ;
3. a resistor;
4. a series resistor and inductor (resistor may be 0Ω);
5. a parallel resistor and capacitor (resistor may be omitted);
6. a series resistor, capacitor, and inductor (resistor may be 0Ω);
7. a parallel resistor (may omit), capacitor, and inductor, with a second resistor in series with the inductor that may be set to zero.

PROGRAM OPERATION

The program prompts for the path to the .ZMA and .FRD files, and then displays a directory of .FRD files; you may select the desired file by number. The file must be no bigger than 20K. The program then presents a directory of .ZMA files so you can select the matching impedance file, which is also limited to 20K and must have the exact same frequency entries as the .FRD file. You must thus be careful to generate both these files with the same IMP setup.

You can then select the frequency range over which to process data and plot/print results, and also select the number of data points to use; 150–200 will provide good plots and limit the number of printed pages. The network is then entered branch by

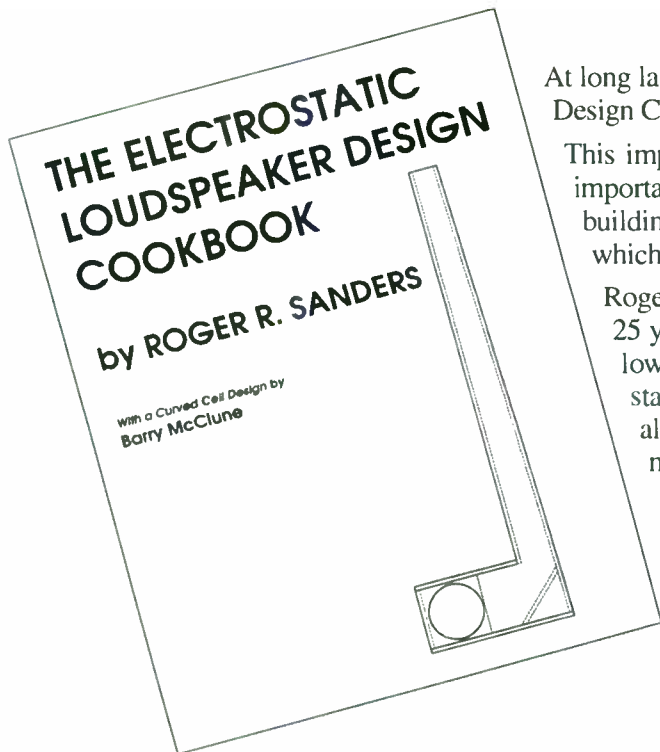


Component values & Branch topologies may be changed from the Main Menu

Any key to Main Menu =

FIGURE 15: Schematic for equalizer network used with tweeter 2.

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branch; finally, the complete network appears on screen. The Main Menu allows you to edit (change values but not topology) the network or modify (total reentry) each branch individually. The program then allows plotting or printing the frequency response and impedance results.

This is not an automatic optimization program; you must manually modify the network. However, the program allows you, from any plot, with a single keypress, to call up any branch for editing or redefining and then return to the updated plot. From a plot, press O for a list of available options.

ROTTEN TWEETERS

Figure 11 shows the IMP frequency response curve of one of a pair of tweeters you wish you had never purchased! They drop dead at about 12kHz and are not flat anywhere. You will never make great tweeters out of them but would like to use them from about 4kHz up to their 12kHz limit in a pair of utility boxes. Figure 12 shows the IMP-generated tweeter input impedance. Note the double peaks at resonance; IMPZobel would never be able to model these jewels!

I briefly played around with IMPPcoEq to obtain the results in Fig. 13. The network

produced a crossover (CO) around 4.5kHz and a response flat from there to about 12kHz. Figure 14 shows the resulting input impedance with the network, which looks acceptable.

Figure 15 shows the passive network, which initially looks very complex. Note, however, that R1, R5, and R7 are omitted. Also, R4 and R8 are really the inductor resistances. The network consists of two capacitors and two inductors—not a high price to use these tweeters considering it is also the CO.

WIMPY BASS

As another example, assume you own small, closed box systems that are great but lack bass extension (Fig. 16). You play them at low level for background sound, so power capability is not a problem. Could you extend the bass without constructing new boxes? The .FRD file for Fig. 16 and the corresponding .ZMA file are fed to IMPPcoEq.

Figure 17 is the result of briefly playing with IMPPcoEq. The bass has been extended from about 60Hz down to about 45Hz. The network response shows that about 7dB of voltage gain is required to accomplish this. How is this possible when we know passive networks can't provide power gain?

Figure 18 shows the Z_{IN} plots from IMPPcoEq and gives the answer to our question. The voltage gain is obtained by the input impedance dipping to about 1Ω. Your amplifier may not like this.

The point is you have learned this sitting at your computer without winding coils and possibly harming the amplifier. I recommend more work with IMPPcoEq to develop a network with less bass extension but a more acceptable input impedance. Figure 19 shows the schematic of the network that was developed, consisting of one each resistor, capacitor, and inductor.

The network branch values are shown on the plots, and a schematic of the network with component values is displayed on screen in VGA mode. If you cannot print the graphic screens, reduce the number of data points and print out a frequency response or impedance list, which will contain the component values for each branch.

IMPEQLZR PROGRAM

Unlike IMPPcoEq, which places a passive network right at the driver/system input, the IMP Equalizer program is designed for developing slow, smooth downslope equalizers that are introduced into the electronics, possibly between the preamp and the amplifier. It is not an equalizer circuit design program but an equalizer function design program. It will show

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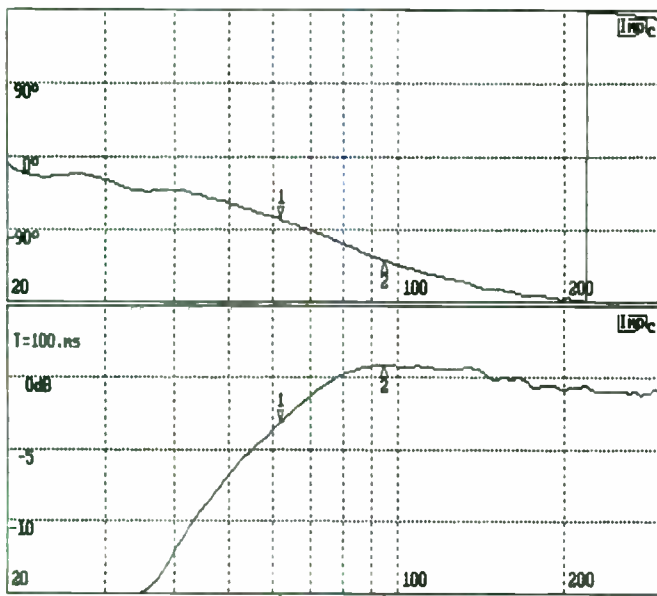
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There are also a number of *Elektor Electronics* books geared to the electronics enthusiast - professional or amateur. These include data books and circuit books, which have proved highly popular. Two new books (published November 1993) are *305 Circuits* and *SMT Projects*. Books, printed-circuit boards, programmed EPROMs and diskettes are available from

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FIGURE 16: IMP frequency response plot of closed box system box 1.

simple passive implementations for the various equalizers developed, but basically helps develop the shape of the equalizer function.

The program takes in a frequency

response .FRD file and applies one or two sections of equalizer to that response. The equalizer slopes are always negative, increasing loss with frequency. Each of the two sections can be -6dB/octave or -12dB/octave , and you can omit the second or both sections. Omitting both will let you plot the IMP .FRD file, which might be use-

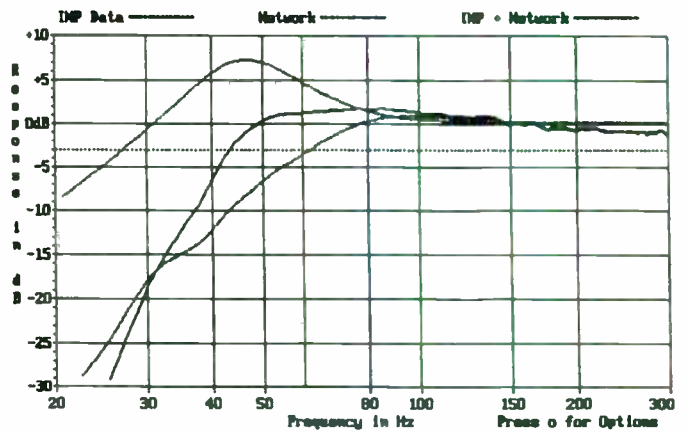


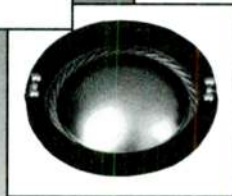
FIGURE 17: IMPPcoEq frequency response plot for box 1 with equalization.

ful if you did not save source data files, since IMP does not recall .FRD files.

The program first enters a path to the .FRD files, then displays a directory so you can select a file by number. The maximum file size is 50K. You can then select a frequency range to process/plot that includes all or a portion of the IMP data range. This program processes all points in the IMP data file that fall within the selected process/plot frequency range. You next enter a normalization frequency that must be inside the .FRD

range. You then enter a normalization frequency that must be inside the .FRD

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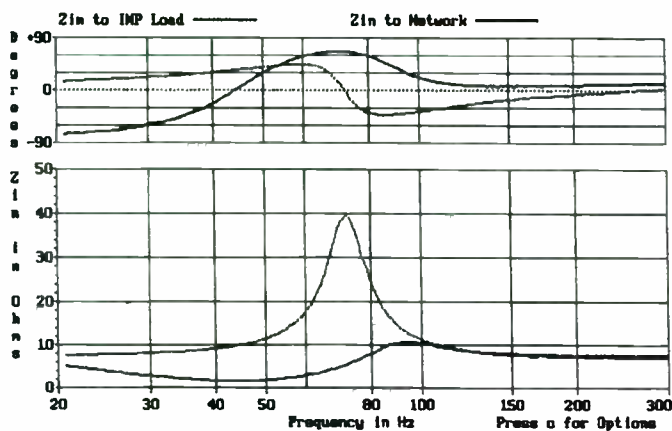


FIGURE 18: IMPPcoEq Z_{IN} plot for box 1 with equalization.

data file frequency range; the actual normalization frequency is displayed, and is the closest frequency in the .FRD file to what you entered.

At this frequency all plots (.FRD data, network, and network plus .FRD data) will be 0dB. This allows all three curves to appear on a single plot. It can affect the looks of the results but does not affect the results themselves. The program also accepts an optional name to appear on the plots/printouts.

You can then define the equalizer network. For the first section you can select: none (omits second section also), -6dB/octave, or -12dB/octave. If you select an equalizer, you must enter the frequency where the equalizer goes from flat to the specified slope (2Hz-5kHz), and also the

the program does not require this. If you set $f_d = f_u$, you obtain a total equalization of 6dB per section (either -6 or -12dB/octave slope). Setting f_u less than f_d can produce equalizations less than 6dB. I kept the allowable ranges for f_d and f_u entry very wide to make the program generally useful. If you select dual -12dB/octave sections with $f_d = 2\text{Hz}$ and $f_u = 20\text{kHz}$, you get over 240dB of equalization!

The program survived this, but be warned if you feed in wild numbers you may get tossed out to DOS! You also must read the equalizer design component values carefully, since resistor and capacitor values can be very large or very small.

IMP EQUALIZER RESPONSES

The program will output tabulated frequency

point where the equalizer breaks back to flat (2Hz-20kHz). All of this program's equalizers start flat, break down at the specified slope, and then return to flat.

The second section has the same options, and has independently defined down break (f_d) and up break (f_u) frequencies. Normally, you would work with f_u well above f_d , but

response data, in addition to the plot. You may preview this on the screen to determine how many points to output. While the program processes all IMP data file points in the selected frequency range, you have the option to output only every Nth point, where N is 1-10. If you select print output, the program will indicate the number of pages in the printout and let you escape it.

Once you develop your equalizer function, the program displays information about it and shows a simple passive implementation with two resistors and one capacitor per -6dB/octave slope. You may enter one resistor value (10Ω to 100kΩ), and the program will provide the capacitor and other resistor value. You can print these text screens with the Shift/Print-Screen keys.

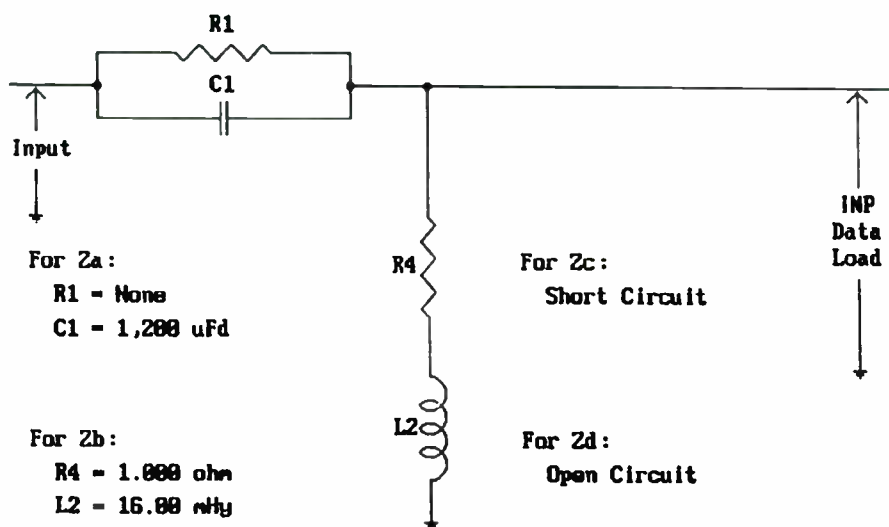
While I used IMPPcoEq to extend the response of a small closed box system, IMPEqlzr is useful when you need a smooth, slow equalization over a wide frequency range, as in correcting the response of an open baffle system. Figure 20 shows the response of a baffle-mounted driver which we will pretend represents the response of your prototype open baffle woofer. Note that this is real-room far-field data, so I included 1/3 octave smoothing to give the basic response shape.

Suppose you experiment with this system, using it up to the 500-600Hz range. The present bass response is down 10dB from the 600Hz response, and thus not worth trying. Figure 21 shows that IMPEqlzr can flatten the general response to make testing the system meaningful. The response is now within ±3dB from about 45-700Hz. The equalizer response shows that 10dB of gain is being used at low frequency, so be careful of the power delivered to the system.

Figure 22 shows the nonnormalized gain figures for the equalizer, along with how to implement the circuit. Note that an additional 13dB of attenuation occurs from 600Hz up until the equalizer is "flat" again. This must be factored into the CO design.

THE IMP-AID DISK

The IMP-Aid disk distributed by Old Colony Sound Lab contains the five functional programs described above. In addition, a menu program (IMPMenu working with XMenu) calls the individual programs for you, and provides additional information about the programs. All the data (.FRD and .ZMA) files used in this article are on the disk. They are straight ASCII files, so you can print or browse through them to see the format used with these file types. Also, a ReadMe.Txt file tells how to install and run the programs. To print this two-page file, log onto the drive containing the IMP-Aid disk and use:



Component values & Branch topologies may be changed from the Main Menu

Any key to Main Menu ■

FIGURE 19: Schematic for box 1 equalizer network.

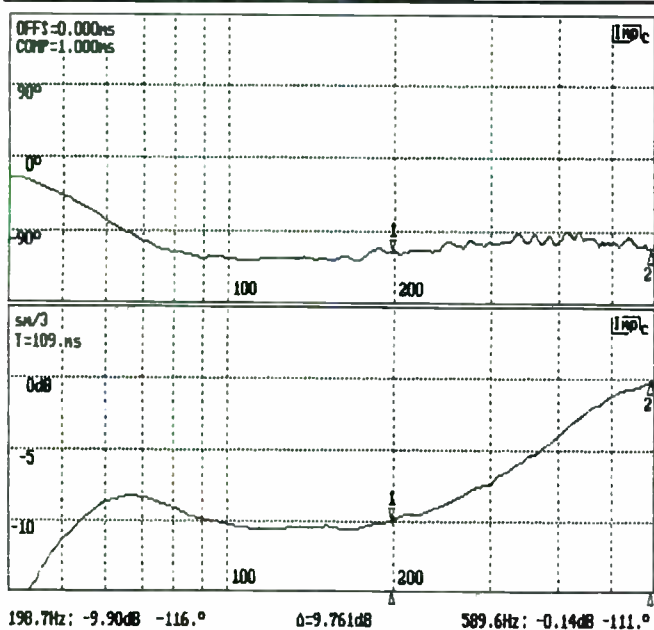


FIGURE 20: IMP frequency response plot for open baffle system box 3.

Type ReadMe.Txt >pm <Enter>

All programs run under DOS on an IBM-compatible computer. The programs should not be DOS version sensitive, but have only been tested with versions 3.21 and 5.0. VGA 640 x 480 pixel 16-color graphics capability is required for all programs (including the Menu system), except for IMPSpace, which always runs in text mode. The programs do not have the ability to print graphic screens. About 280K available memory is required. Some of the programs are floating-point computation intensive, so a fast computer with coprocessor is beneficial.

Since I developed these programs over a long period, they use a variety of interfaces. Details of these differences are included in the Information section of the Menu system. All interfaces are text-based; however, the Menu system and IMPRoom have mouse support.

The programs are supplied on a single high-density 5.25" (1.2M) or a single high-density 3.5" (1.44M) disk. To reduce the total program size, three of the programs are self-extracting compressed .Exe files. No special action is required on your part to load and run these files, which act just like ordinary .Exe files but require more memory than their file size indicates.

FILE SIZE REDUCTION

It is possible with IMP and very easy with Liberty Audiosuite to generate .ZMA or .FRD files that are too large for the post-processing programs described above. To solve

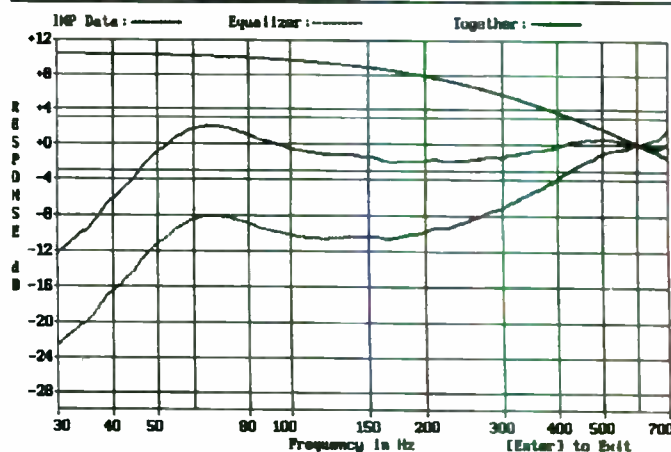


FIGURE 21: IMPEqlzr frequency response plot for box 3 with equalization.

this problem the utility program Reduce.Exe is included on the IMP-Aid disk. This program must be

called directly, as it is not included in the menu system; the program requires VGA graphics capability and offers mouse support. See the program information section of Reduce.Exe for details of its oper-

ation. In general it takes every Nth line from the original file and writes them to a new file, where N is set to provide the file size you desire. The first and last lines of the original file are always retained so the frequency span of the file is not reduced. ReadMe.Txt on the IMP-Aid disk discusses running Reduce.Exe.

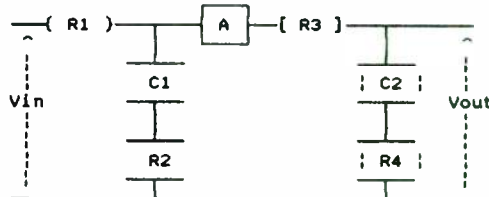
Two Section Equalizer Performance

0dB	Slope	Performance for Total Equalizer	
First Section	-6dB/oct	fd1 = 300.00 Hz	Real Gain = -4.68 dB
	fu1	fu1 = 1600.00 Hz	Real Gain = -18.98 dB
	fd2	fd2 = 900.00 Hz	Real Gain = -14.32 dB
Second Section	-6dB/oct	fu2 = 1200.00 Hz	Real Gain = -16.87 dB
	fu2	fg = 600.00 Hz	Real Gain = -10.41 dB
		Infinite frequency	Real Gain = -23.39 dB

M to Main Menu, nul to Design page:

Two Section Equalizer Design

First Section: -6dB/oct Second Section: -6dB/oct



Note: [A] equals a Unity gain Buffer Amp.

R1 = 53.33 Kohms	R2 = 10.00 Kohms	C1 = 0.009947 uFd
R3 = 13.33 Kohms	R4 = 10.00 Kohms	C2 = 0.013263 uFd

Enter Value of R2 (10 to 100000), nul to Exit - Ohms:

FIGURE 22: Design and performance of equalizer for box 3.

Wayland's Wood World

SHAPED HOLE TECHNIQUES

By Bob Wayland

Mounting a speaker needn't involve free-cutting saber saws and sanding drums. The results with these, all too often, are disappointing. There are easier ways to produce far more acceptable, professional-looking mountings. This column explains how to make circular holes for both small and large speakers.

FOR SMALL HOLES

The mounting holes for smaller speakers less than 5-6" in diameter are simpler to make than for larger speakers, due to a number of well-designed hole cutters (Photos 1, 2, and 3). The expansion bits are used in a hand brace to give you great control. To improve your hole cutting, clamp a backup board behind the hole you're drilling. This helps in two ways: the backup board provides a base into which the screw tip of the expansion bit can continue, and prevents tearout of the board on the back side of the drilled hole.



PHOTO 3: Expansion bits for use with a brace. The maximum radius is limited, usually to less than 2".

Of course, you can use a hand power drill with hole saws and Forstner bits, but this requires a very steady hand and a keen eye to keep the drill perpendicular. If you don't, the saw will produce excess heat and result

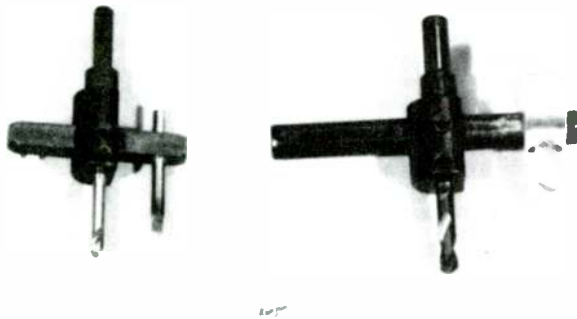


PHOTO 1: Adjustable fly circle cutters for drill presses.

in a cockeyed hole. Use a drill press whenever possible.

Forstner bits produce one of the cleanest holes and have the distinct capability of making flat-bottomed holes (Photo 4). They are available in diameters up to 3-1/8" from most woodworking and mail-order supply houses. Use bits larger than 2" only in a drill press.

However, they are not cheap; expect to pay up to \$25 for a 2" bit. You might also consider a spur bit, which must be used only at 90° to the surface. They usually cost about 50% more than Forstner bits.

One of the most useful cutters for 2-6" holes is the fly cutter (Photo 1). (Most people refer to them as circle hole cutters.) This is simply a cutting bit attached to the end of a movable arm, which is anchored in a mandrel into which a bit is attached to provide a well-defined center for the rotation. Because tremendous forces act on the cutting tip and arm, be sure to clamp your work to the drill press table and keep your hands away.

I prefer a fly cutter with a counterweight at the end of the arm to provide damping and increased angular momentum (Photo 5). You will usually have better luck if you use a low-to-medium speed (500-1,000 rpm).

The accumulation of cutting debris can cause heat, so raise the bit often to help clear the debris and cool the bit. Again, a backup board is helpful in controlling tearout on the back side of the hole. Although you can cut

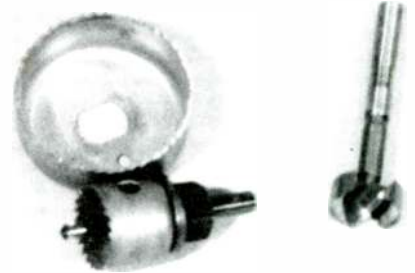


PHOTO 2: Fixed radius hole cutters include hole saws (left) and Forstner bits (right).

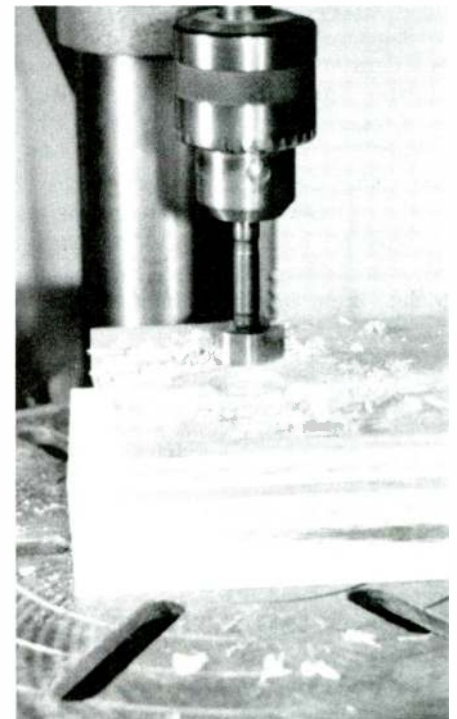


PHOTO 4: Forstner bit to drill a flat-bottomed hole. Note the backup board under the board being drilled.



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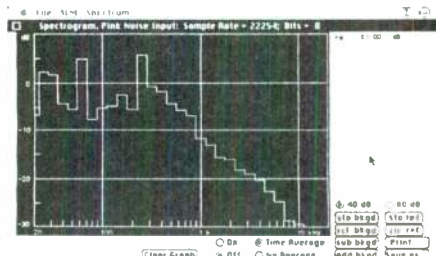
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Elaborate English/German/French 39-page manual features graphs, diagrams, photos. Japan, 1988, 53:49.

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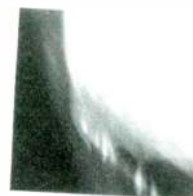
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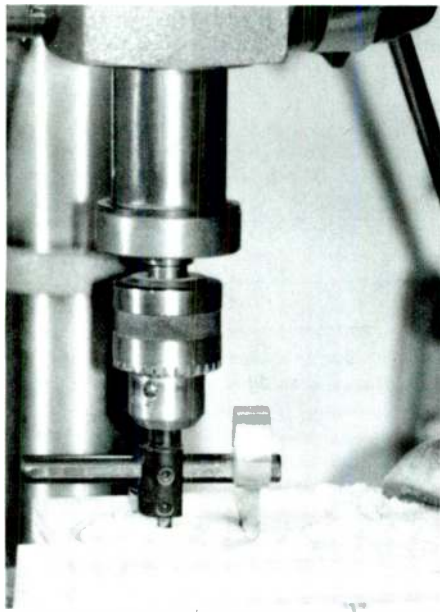


PHOTO 5: A counterbalanced fly cutter to cut a hole. This helps cut down on chatter at the cutting bit. (Note the clamp on the right side.)



PHOTO 6: Cutting the adjustment slot for the router jig, with the mounting holes and the center hole marked and drilled out. The mounting screw holes have been countersunk so the screws will not scratch the surface of your piece.

holes bigger than 5" with a fly cutter, it is not a good idea. Also, if you plan to remain whole into old age, use a drill press, never a hand drill!

FOR LARGE HOLES

Larger midrange and woofer speakers require larger holes, some up to 15". Fly cutters are not the answer. If you have a saber saw or router, you can easily make a simple jig that really works. If you have a fence for your router, you can perform professional-quality

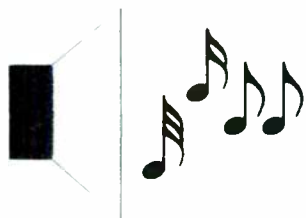
work. (More on this later.)

You can make the jig for your saber saw or router from a piece of scrap 1/4" material (plywood works fine). The following techniques apply to both a saber saw or router, although I don't like saber saws for this purpose. They produce serviceable holes, but they do not work as well as a router.

If you're using a router, remove the faceplate from the bottom. On a piece of 1/4" plywood, 18-24" long and at least as wide as the faceplate, transfer the position of the holes

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PHOTO 7: Screw, washer, lock washer, and T-nut assembled into the slot in the plywood jig slot.

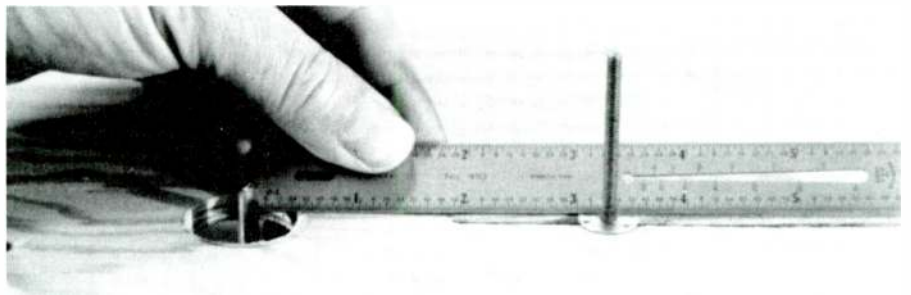



PHOTO 8: Adjusting the circle cutting jig using a ruler to set the radius. The far side of the router bit determines the radius.

(Photo 6) and drill them out, countersinking the mounting holes to accept the original mounting screws. Draw a line parallel to one edge of the plywood, using the center of the center hole as a guide.



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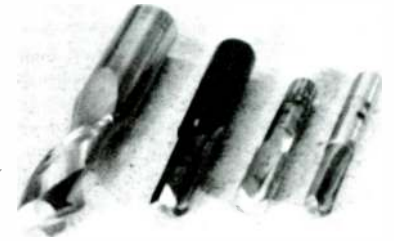


PHOTO 9: Straight bits for routing. My favorite is the spiral bit on the left.

Measure the outer diameter of the threaded portion of a 6-32 x 1/4" T-nut. This is the width of your cut, which should not extend to the center hole or through the end of the board opposite the center hole.

One way to make the cut is on a table saw with the spreader removed (Photo 6). Align the plywood so the slot will start inside the piece about an inch from the edge. Gently lower the plywood onto the running saw blade while firmly holding the back edge in place. (I removed the blade guard to take this photo. Use the blade guard, whenever possible, while working on a table saw.)

Make the cut, stopping before the center hole. If the width of the cut is not enough to accept the shank of the T-nut, you'll need to make a second cut.

Put a washer and lock washer on a 1/4" 6-32 bolt, and screw it into the T-nut with the washers on top of the plywood (Photo 7). Tightening the screw locks the assembly in place to act as a pivot around which to cut a circle. You can adjust the screw pivot of the assembled system to provide a specified radius (Photo 8).

You will need a straight bit for your router (Photo 9). I prefer the spiral-flute solid carbide bit, which costs about \$15-\$25 from Grizzly Imports. It will keep the cut clear of chips and produce clean, sharp cuts. If possible, use 1/2" shanks for added stability.

Next, mark the location for the hole



PHOTO 10: Laying out the circle.



PHOTO 11: Using the jig and router.

(Photo 10) and drill a center hole big enough to just accept the screw pivot. Now, simply rotate the jig about the pivot (Photo 11), cutting out a circle. For best results with a router, don't try to cut all the way through on the first pass; cut about an eighth of an inch at a time. If you use a backup board, the pivot screw should extend down into it, which keeps the cutout stock in place as you make the final cut.

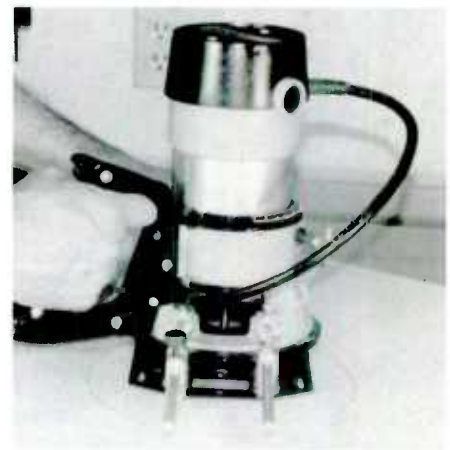


PHOTO 12: Using a modified router fence.

USING ROUTER FENCES

You can use this basic idea to modify a router fence (also called a router edge guide) to make holes. The fence replaces the plywood jig. Some manufacturers have built this into their fences, but other models require slight modifications.

The guide on the fence is usually a stamped-out piece of sheet metal with a flat surface flush with the base of the

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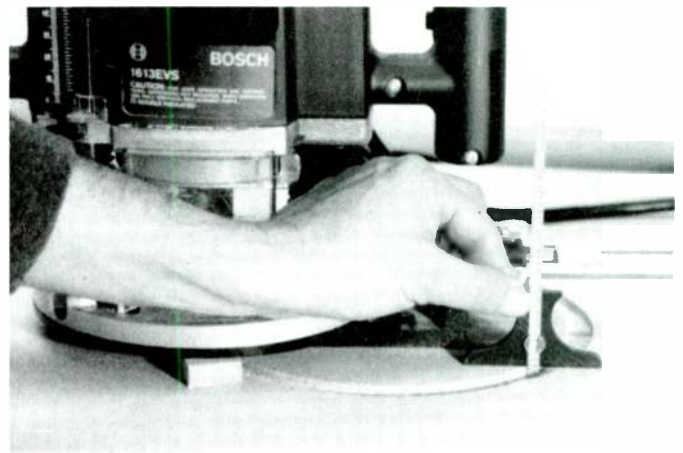
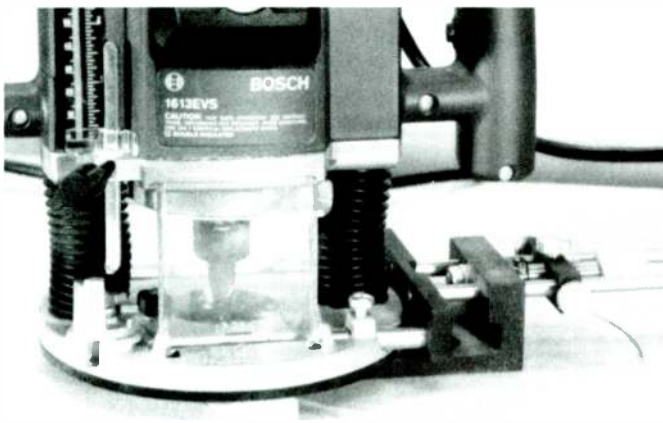


PHOTO 13: Micro Fence attached to a plunge router.

PHOTO 14: Checking the depth of cut for a recessed ring that doesn't go through the board.

router. Draw a center line between the two rods extending from the base. If there is not a hole on this line, you will need to make one. You can mount a screw through this hole to use as the pivot. Follow the same procedure as described above for the homemade jig, except substitute the fence for the jig (*Photo 12*).

If your router fence extends below the plane of the base, you will need to make a wooden spacer that mounts on the bottom

of the router—just thick enough to bring the bottom of the fence and the spacer bottom into the same plane. Instead of the screws as pivots, use spacers and standoffs. As audiophiles we normally use them to separate or mount electronic boards, but the round ones make perfect pivots. Those with a 8–32 threaded stud are good choices. Be sure the spacer is long enough to penetrate your board and at least 1/4" into the backup board.

STEP BY STEP

All of the above methods produce perfectly acceptable holes for mounting your speakers. Another, rather expensive, device you can purchase to make quality holes is the Micro Fence (*Photo 13*). This router fence (\$98 from Micro Fence) is an after-market addition that is precision made for professional results. It is custom-manufactured to fit individual routers. You can set the position of the fence with a precision of 0.001".

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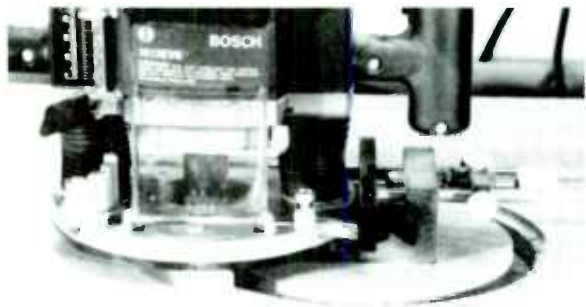


PHOTO 15: Making the final through cut on a two-stepped speaker mounting hole.



PHOTO 16: Final configuration of the two-step mounting hole.



PHOTO 17: Speaker mounted in the board. Note how closely the speaker fits into the hole. This set of cuts was made on particleboard; MDF produces even better results.

Before you can use it to cut circles, you must modify the main body of the fence. As above, find the center line on the plate between the two rods extending from the

router base. Drill a hole with a #59 drill about 1" in from the surface nearest to the router on the center line. Tap the hole for an 8-32 thread. Then attach a circular spacer with an 8-32 stud by screwing it into the hole.


The shoulder of the spacer will help provide support of the spacer that now becomes your pivot for cutting circles. The spacer should be long enough to extend into the backup board. Because the main body of the fence will be flush with the top of the board

you are cutting, you will need to attach a block of wood to the base of the router to keep the base parallel to the board. *Photo 13* shows this spacing block under the base of the router. Be sure to mount the board outside the circle you are cutting.

Normally I prefer to mount the speaker so the front of the mounting flange is flush with the mounting board. This requires a two-or-more-stepped hole, with one recessed ring that doesn't go completely through the board. Of course, if the speaker frame is more complex, you may need to cut more than two steps. Measure the radii and the depths needed for each ring.

Assuming that the shallowest ring is on the outside, set the radius 1/32" less than you need. Make a very shallow test cut. Measure the radius and then, using the micro adjust feature, set the fence to meet your exact requirements. Start cutting the ring, stopping often to measure the depth (*Photo 14*).

Move the fence to the next radius. Correct as above after you make a very shallow test cut. Continue this process until you have made the final through cut (*Photo 15*). Drill the holes for attaching the speaker and apply any necessary sealant. The speaker should fit snugly into the finished hole (*Photo 16*). The results of your efforts should be enough to make any speaker builder proud (*Photo 17*).

You can use the technique outlined above for producing stepped holes with either the jig or the standard fences. My next column describes how to make mountings for irregularly shaped speaker flanges. 

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CD Review

MY DISC FROM SHEFFIELD LAB

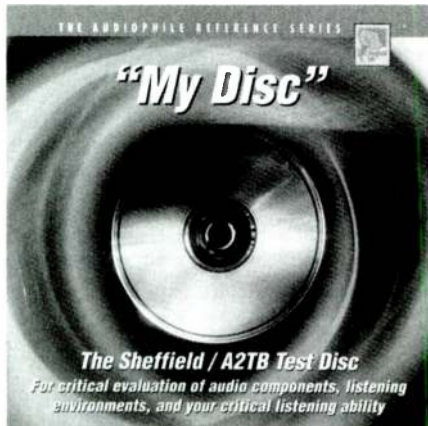
By Dan Ferguson

My Disc, Sheffield Lab Recordings, 1046 Washington St., Raleigh, NC 27605, (919) 829-1154, FAX (919) 829-0047. [Available as #CDSH3 for \$29.95 plus \$3 S/H in US from Old Colony Sound Lab, PO Box 243, Peterborough, NH 03458-0243, (603) 924-6371, FAX (603) 924-9467.]

One of the real frustrations for audio hobbyists and sound enthusiasts is the inability to measure objectively the performance of our handiwork. Few hobbyists have access to expensive test equipment and more often than not final evaluation of a construction project becomes very subjective. In addition, we have difficulty evaluating our own creations without bias.

The need for assistance was recognized 30 or 40 years ago, and from time to time test records and CDs have appeared on the market as answers to the problem. A new CD, *My Disc* from Sheffield Lab, is being billed as slanted toward the needs of auto sound. However, I find it suitable for any audio application.

As the third in Sheffield Lab's "Audiophile Reference Series," this disc was a joint effort with auto sound authority Richard Clark of Autosound 2000. The disc's ambitious intentions are accurately described



by Mr. Clark in the instruction booklet introduction. *My Disc* is intended to be a singularly complete test medium for everything from the listener's ears and ability to distinguish nuances in program material to the equipment and listening environment.

To accomplish such a comprehensive agenda obviously requires a large amount of information, which *My Disc* has: 74 minutes worth on 86 tracks. While this may seem a little overwhelming at first, once the tracks are grouped by category the disc's uses become clear. In my opinion, the tracks could be better organized. To aid our discussion I have

regrouped them into Music, Imaging, and Technical Tracks. The first two categories are simply for listening, while some test equipment is needed for most of the third.

MUSIC TRACKS

Tracks 1-6 are examples of various types of recorded music (pop, jazz, and classical) selected for their exceptional ability to demonstrate a playback system's detail and imaging. The instruction book points out the features on which the listener should focus to make the evaluation.

I found these tracks to be both enlightening and disappointing: the former from an "ear training" standpoint and the latter from a system builder's standpoint. *My Disc* enabled me to see clearly that the imaging of my home and car systems was less than ideal in some areas. System analysis aside, these tracks are very enjoyable listening.

IMAGING TRACKS

Track 19 is simply a recording of a person counting, and serves as a "reality check."

Tracks 20 and 21 identify right and left channels.

Track 22 is a channel-to-channel phase check. As a person speaks, the right and left channels are switched in and out of phase

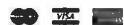


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for comparison. The test makes actual phase conditions easily recognizable and would be extremely valuable in auto sound applications where wiring errors are so easily made. I've actually had to remove the speakers in a client's car with an imaging problem just to inspect the wiring and verify polarity.

Tracks 25 and 26 consist of three people speaking and describing their locations as right, left, and center. I suspect this series would be a good test to determine whether you are overdoing a center channel (if you're fortunate enough to have one).

Tracks 27-32 contain specially

recorded musical passages which test lateral imaging by shifting from normal stereo to mono, panned left, back to center, right, and finally back to center. This gave both of my systems a real workout, and enabled me to easily detect some minor weaknesses in my truck's front soundstage.

TESTING PARAMETERS

The remainder of the disc is devoted to a variety of technical tracks capable of wringing out your entire playback system, from your CD player's digital-to-analog converters to the overall phasing and response. But before

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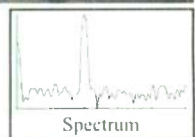
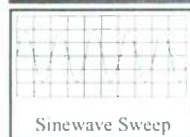
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Reader Service #23

Speaker Builder 4/95 49

we proceed, let's discuss some practical aspects of testing with *My Disc*.

In order to use many of the technical sections with any real effectiveness you will need an accurate microphone such as Old Colony's Mitey Mike or a sound level meter (SLM) like those sold by Radio Shack. The SLM has the advantage of being self-contained with everything you need in one compact package. The disadvantage is, at a retail price of \$29.95, it has obvious accuracy limitations. [The analog version may be improved by adapting it for use with Mitey Mike. See C.L.P. Carrington, "Converting Radio Shack's SLM to Millivolt Use," SB 4/94, p. 12.—Ed.]

On the other hand, the Mitey Mike has terrific accuracy but you need some way to read its output. Here, an accurate, "true RMS" meter is required. Most of them are quite expensive and yet have limited bandwidth, rendering them incapable of covering the entire audio spectrum. All of this can be somewhat perplexing.

I decided to make some trial measurements with my Radio Shack SLM and compare the results with those of my DOD real-time analyzer. I got surprisingly good agreement between 20Hz–12.5kHz, at which point the SLM response begins to fall off. The calibration curve supplied with my SLM shows

the meter beginning to fall off in that region and being down 20dB at 20kHz. It appears to be down approximately 5dB at 16kHz and 11dB at 20kHz, which is a flatter response than expected.

I would say that my Radio Shack SLM will give good results between 20Hz–12.5kHz. Beyond that, the calibration is somewhat unknown. Armed only with *My Disc*, a \$29.95 SLM, some time, and a great deal of patience, you should be able to tune your system reasonably well within that range.

TECHNICAL TRACKS

Tracks 7 and 8 are reference level 1kHz sine waves at -20dB f_s and 0dB f_s , respectively. The liner notes indicate that these demonstrate the 20dB headroom available on CDs, and that 0dB is the highest recordable signal level in that medium. I regard these tracks as being in the "general information" category.

Tracks 9–18 are ten warble tones (one per track) which are one octave wide and centered at 20, 62, 125, 250, 500, 2.5k, 5k, 10k, 15k, and 19kHz. Most of these are located near some popular graphic equalizer center frequencies, which should aid in system tuning. As most speaker builders know, warble tones are used to minimize the effect of standing waves in an attempt to remove the response of the listening space from the mea-

surements. I found this section extremely easy to use and an excellent starting point for tuning a graphic equalizer.

Track 23 consists of polarity pulses which are intended to be used with a polarity checking instrument.

Track 24 is a clipped 1kHz sine wave intended for use with an oscilloscope as a signal tracer.

Tracks 33–42 test the linearity and accuracy of your CD player's A/D converter by playing a reference musical passage at successively lower levels. Linear reproduction of low-level signals has always been the most difficult task for digital systems. So, if the music sounds clean through each iteration, your A/D converter is working well. My Sony with 4x oversampling began to hiss audibly at -40dB. At successively lower signal levels, the hiss increased proportionately. This was certainly the first I had heard anything other than music from my CD player. The problem is I don't know whether a threshold of -40dB is good or bad performance.

Track 43 is "correlated" pink noise (identical signals to both channels). Pink noise is composed of all frequencies, at random, with equal energy (not amplitude) contained in each octave. In my opinion this is the most valuable track, but only if you have access to

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a real-time analyzer (RTA). I own an inexpensive (\$300) DOD RTA-II, and it is a real joy to be able to measure the results of my efforts with some degree of accuracy. Using this track and the RTA produced a delightful-sounding tuning of the new ten-band Jensen graphic equalizer in my truck. A friend described the sound as smooth and textured, and likened it to sipping old, expensive whiskey.

Incidentally, these pink noise tracks last only about a minute. So you will need to program your CD player for that single track in repeat mode.

Track 44 consists of "uncorrelated" pink noise, which means the noise in each channel was recorded independently. While I have not encountered this before, the instruction book states it can aid in tuning a system where the listener is not equidistant from each speaker (like in a car). The recommendation is to tune with both correlated and uncorrelated noise, then compromise between the two. From a practical standpoint I found this somewhat tedious; however, if a person were building an auto sound system for competition purposes, it may provide an edge.

Tracks 45 and 46 demonstrate the logarithmic concept of decibels through a series of volume changes.

Tracks 47-56 contain 30 bands of pink noise (1/3 octave each) from 25Hz to 20kHz. If you wish to measure the response of your system and don't own an RTA, these tracks make *My Disc* invaluable. As discussed previously, you will need some type of sound level measurement equipment to use this. For pure convenience, the SLM is hard to beat. Even so, this is a tedious process that requires quite a bit of patience.

To equalize your system, draw a graph with 30 points on the "X" axis and about 20 divisions on the "Y" axis. As the disk cycles through the bands, plot the sound levels on the graph. Adjust your equalizer settings accordingly and repeat the process. With a little perseverance, you can duplicate the results obtained with an RTA.

Even if your system does not incorporate a graphic equalizer, these tracks can serve other uses. One would be to determine the best settings for individual driver L-pads or bass and treble controls.

Track 57 is a continuously swept warble tone from 20Hz to 20kHz, which "subjectively" evaluates your system's overall smoothness. In addition, it's another outstanding tool to provide a final check with your SLM after Tracks 47-56.

Track 58 is similar to the preceding track, except the frequency range is 100-20Hz for subwoofer evaluation.

Track 59 contains low-distortion sine waves at 90 individual frequencies from 10-99Hz. The instructions suggest using them to find the resonant frequency of low-frequency drivers, and, consequently, as a signal generator. This would work well if the frequencies were in 0.1Hz increments; as it is, they are of limited value. However, if a CD were developed with finer frequency increments, I could envision measuring a driver's Thiele/Small parameters simply by plugging into a CD player's headphone jack!

Tracks 60-62 are a low-distortion swept sine wave from 20Hz to 20kHz, primarily for evaluating electronic components.

Track 63 is an accurate musical tone (A at 440Hz) which can serve as a reference to tune musical instruments, similar to a tuning fork.

If this section were expanded to all 88 piano tones, you could conceivably tune anything accurately.

Tracks 64-66 are tone bursts at different frequencies, which test the transient response of various components. A microphone and

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oscilloscope are needed to make the best use of these tracks.

Tracks 67 and 68 are square waves at 100 and 1kHz, respectively. A good application for these tracks is "time alignment" of individual drivers. Again, a mike and oscilloscope are required.

Tracks 69 and 70 demonstrate the concept of dynamic compression of conventional recordings prior to CDs. They can also be used to check the accuracy of tape deck recording level meters. Incidentally, if your meters are accurate at all frequencies, you should be able to employ them with a microphone as a SLM.

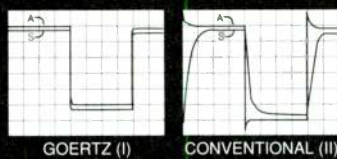
Tracks 71-76 are a sine wave, initially with a very low 0.004% distortion and increasing in steps to 10%. These tracks check the calibration of a distortion analyzer (very expensive) or simply function as ear training.

Tracks 77-82 are similar to the preceding ones, except the signal source is music. The main application here is ear training and a test of your ability to distinguish distortion. It is quite surprising to learn the extent to which music can mask the distortion levels that were so easily detected in the previous sine wave test.

Tracks 83 and 84 demonstrate the concept

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—Clark Johnson, Positive Feedback, October 1994

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of "group delay." I found the effects to be very subtle.

Tracks 85 and 86 are the sounds heard on pit row at the Indianapolis Motor Speedway. According to the instructions, they are best appreciated by listening through headphones. I must agree. These are nice novelties that add to the overall comprehensive nature of *My Disc* but are of lesser value in system testing.

IN CLOSING

With the number of test CDs and records on the market, someone seeing an ad for *My Disc* might be tempted to say, "Oh no, not

another one!" This one, however, just may be different and better. It definitely meets its ambitious objectives, and I heartily recommend it.

MANUFACTURER'S RESPONSE

Thank you very much for your very thorough review of My Disc. This is truly a project designed for anyone who wishes to gain a better understanding of their system's capabilities. I would like to mention only a couple of points.

One of the reasons My Disc is selling so well, despite the ever-growing population of test software available today, is that it was

designed for the listener to make more accurate judgments about the performance of a given playback system. But the unique feature about this newest member in Sheffield's catalog of "Test Tools" is its focus on allowing the listener to evaluate his/her own critical-listening ability, not just the performance capabilities of the audio system in a given environment. We feel that educating the listener is the most important key in helping an individual to understand, develop, and improve a sound system. This is dependent upon the listener having some idea about his/her own hearing capabilities.

The other point which Mr. Ferguson certainly makes clear, but that I would like to emphasize even further, is, despite the fact that My Disc was created in cooperation with Richard Clark (the highly distinguished engineer from Autosound 2000), this disc is equally useful as a home-audio test product as it is in the car. This might be an important point to make, especially considering that although Ferguson tried out My Disc in his car, I would imagine that the bulk of the folks who subscribe to Speaker Builder are home-audio enthusiasts.

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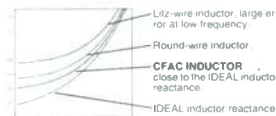
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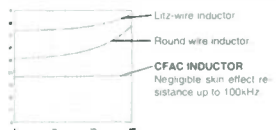
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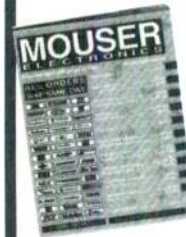


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CORRECT TEST RESULTS

The incorrect Fig. 7 was included in "Box Models: Benson Versus Small" (*SB* 3/95, p. 14). The correct curves are shown in *Fig. 1*.

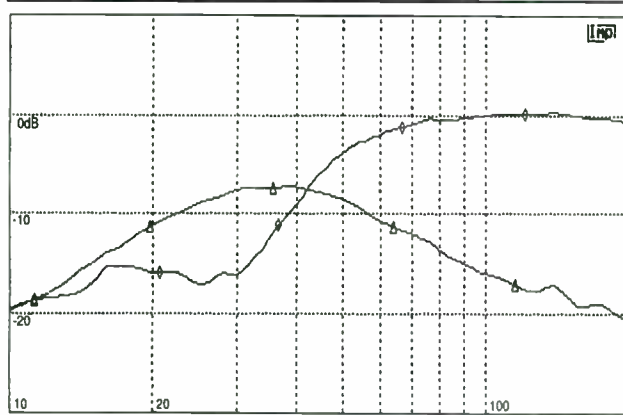
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In a recent letter to "SB Mailbox" (*SB* 2/95, p. 60), I mentioned the source for Acoustic Research loudspeaker replacement parts, but included the incorrect telephone number. The following is the complete, corrected information:
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△ memory trace:
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FIGURE 1: Figure 7 for "Benson Versus Small" article.

REMEMBRANCE

The passing of my old friend Raymond Cooke OBE was more than a little personally distressing, since once again, the band of my generation of audio enthusiasts is depleted by someone we can ill afford to lose. They are all a special breed, whose interests are entirely music-driven. This is in contrast to what I see as a disturbing trend, an excessive interest in the mechanics and insufficient regard for the end product which, after all, surely is what matters.

It was a search over many years for a remarkable performance of the *Dream of Gerontius* that led me back to Raymond, whose love of Elgar was well-known, and the composer's greatest achievement. It was characteristic of the man that he called me from his hospital bed



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to confirm that he had the recording, and we discussed ways of getting it to me. Yes, it was that precious. Sadly, there were many problems to be overcome first.

It was Betamax, made on a PCMF1 to the NTSC standard, and there was a delay before I had a pair of DATs that I could play and evaluate. It had not weathered the years well, requiring extensive editing and repair work.

Despite a solid week's work on the computer, finishing it appropriately on Easter Sunday, it saddened me I could not have completed it before his death a few days earlier. So he never heard this remarkable recording, now restored to pristine glory. Fortunately, his love of the work was shared with Jennie, his wife; so I was able to return it for the pleasure I hope it gave her.

Reg Williamson
Kidsgrove, Staffs, UK ST7 4DE

HORN DESIGN

Thanks for the excellent article on the show horn ("The Show Horn," SB 2/90, p. 10). I've always been in awe of the power of a bass horn, but had no idea how to design

one. I now believe it is within my reach. All the formulas work, but I still have a few questions.

Through what point do you measure the length of a horn, especially around a turn? Does the horn expand through a turn? How do you determine an appropriate size for a throat area?

I've also read Rick Steiners' article ("A Back-Loaded Wall Horn Speaker," SB 4/91, p. 10). Can you tell me the difference between your free space horn and a wall horn? Are the alpha, M-factor, mouth-area, and horn-expansion curve equations the same as for the free space horn?

M. Brewer
Alton, IL 62002

Contributing Editor Bruce Edgar responds:

Thank you for your compliments on the show horn article. I usually measure the length of the horn along the center of the duct. This procedure also applies to turns. Afterwards, I usually draw in the diagonal reflectors.

The horn expands through a turn. The entrance and exit areas conform to the horn expansion through the turn, but don't try to


have the horn expand exactly in the turn. The diagonal reflector prevents an exact expansion through the turn.

The optimum throat formula in my article gives the best results. I only deviate from the formula when the driver has excellent sensitivity (i.e., 99 dB). A good rule of thumb is a throat size that is 50% of the driver area.

A free space horn has a mouth size given by the following formula

$$A_M = \frac{1}{4\pi} (c/f)^2$$

where c = speed of sound, f = flare frequency, and A_M = mouth area that is one-fourth of a free space horn. The acoustical images behind the wall and the floor effectively multiply the area of the mouth by a factor of four.

The alpha, M-factor, and throat-area calculations are independent of the mouth area. You select your driver based upon bandwidth (mass rolloff frequency and flare frequency) and throat area. Run out the horn expansion based upon mouth-boundary requirements. Determine the folding scheme, and draw out the horn expansion. It's that simple. 

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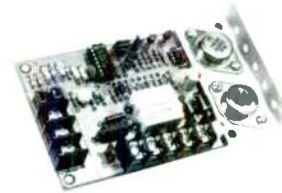
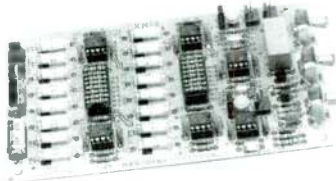
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Reader Service #73

Simpline Sidewinder

from page 9

EPILOGUE

I have discovered another speaker which works well in the Sidewinder: the 8" Parts Express treated paper cone woofer (#295-240). This unit is more rugged in appearance than the Radio Shack 40-1024, and it has a vented pole piece. (It's also much cheaper.) As it is a drop-in replacement, all the holes match. Both speakers perform well in the Sidewinder.

In balancing the woofer with the Simplines, I found I needed no attenuation on the high-pass sections. I set the control on my Parasound HCA-800 somewhere between the 2:30 and 3:00 positions. If the midrange gets muddled with the woofer in place, you have it turned up too high. Another clue that the woofer is set too high is a fairly prominent bass line that lacks the natural transients.

Maintaining midrange clarity and developing the proper transient feeling are to me the most straightforward and easiest ways to balance (any) satellite to a woofer. The settings I have indicated will probably be different for you, unless you are using the same amplifiers; however, they may serve as a starting point. Your ears will dictate the correct position.

This system works well with a single summed woofer set between the Simplines, and may save you time and money. If you decide to make two Sidewinders, placing the input terminals in the lower center of the side panel (instead of the lower rear corner, as shown in the drawings) will enable you to use either one as a right or left channel. It would save the effort of creating a mirror-image pair to maintain the symmetry of the stereo system.

I think you should wind your sides in opposite directions, on alternate days, to keep your symmetrical integrity. You'll like what you hear.

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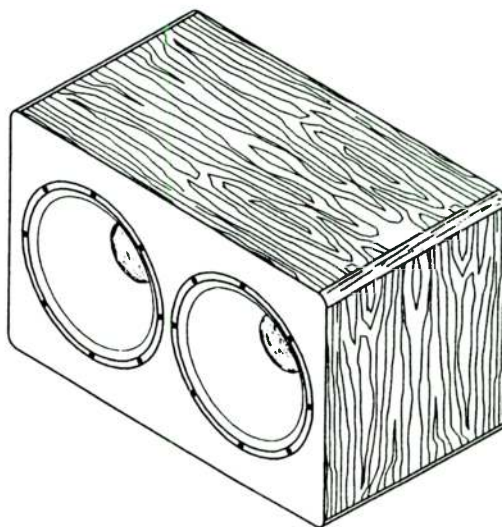
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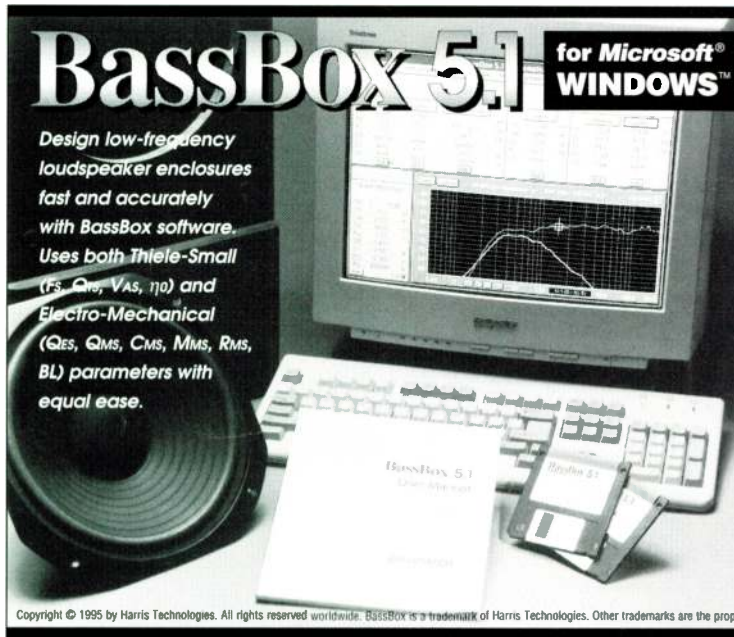
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REFERENCES

1. Olive, Schuck, Sally, and Bonneville, "The Effects of Loudspeaker Placement on Listener Preference Ratings," *JAES*, Vol. 42 (Sept. 1994): 651.
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3. G.R. Koonce, personal correspondence, October 1994.
4. D. Queen, "The Effect of Loudspeaker Radiation Patterns on Stereo Imaging and Clarity," reprinted in *Loudspeakers*, Vol. 2 (AES Anthology): 69.



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The Boston Audio Society
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E. Brad Meyer, phone/FAX (617) 259-9684,
E-mail 72365.75@compuserve.com

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Southeastern Michigan Woofer and Tweeter Marching Society
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