

electronics

TODAY

INTERNATIONAL

FREE
CATALOGUE
from
EDGE
ELECTRIX



Four basic games
Two ball speeds
Two or four angles • Two bat sizes
On-screen scoring • Sound effects

TWO GAME

PRINT-OUT
A MINI-MAGAZINE
FOR ALL MICRO-
COMPUTER-
USERS

Electronics Today International
is Australian owned & produced

Registered for posting as a publication
Category C.

Well stacked in front



The new range of JVC front-loading cassettes is here. And if you think that's the only change, you're highly mistaken. Because, as usual, JVC brings in the range with a few unique additions which are going to make you think twice about any other brand.

For a start, the JVC ANRS sound reduction system is incorporated throughout, to make hi fi recording and playback as free of hiss as possible. And in some cases, even improving the dynamic range of normal cassettes.

Another exclusive is the JVC Sen-alloy head, and believe it or not, it offers you the clearest sound and longest wearing lifespan of any head available; originally designed solely for

professional use, this head is now incorporated in JVC cassette decks CD-S200 and CD-1970.

And yet another first: JVC is the only manufacturer to provide decks with 5 LED peak-level indicators so that your recordings are perfect at all times. These are featured on models CD-1920 and CD-S200.

Loading is, of course, simplified. The special compartment is air-damped and removable for uncramped head maintenance.

The JVC famous range of top-loaders is still available, offering you the very highest quality. All things considered, there is no other consideration.



JVC

the right choice

For details on JVC Hi Fi Equipment, write to: JVC Advisory Service, P.O. Box 49, Kensington, N.S.W. 2033.

electronics TODAY



A MODERN MAGAZINES PUBLICATION

INTERNATIONAL

NOVEMBER 1976, Vol. 6 No. 11

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Electronics Today International is Australian owned and produced. It is published both in Australia and Britain and is the fastest growing electronics magazine in each country.

SPECIAL OFFER
'Sleeping Dog'
Car Alarm . . .
see page 35

NEW SECTION FOR
COMPUTER
ENTHUSIASTS
Starts on page 85

COVER: Build one of the best video games around — and whatever your level of skill you will be able to select a game to match. If you think you are really good try playing our soccer game — with a high speed ball, small men, and four-angle passing. The project starts on page 44.

* Recommended retail price only

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Another in our 'Project Electronics' series

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Extracts from an address by Mr. E. Nakamichi, President Nakamichi Research Inc. at a recent Seminar in Sydney for Nakamichi dealers.

"TDK Super Avilyn Cassettes are recommended for use with all Nakamichi tape decks. Before leaving our factory, all Nakamichi equipment has bias voltages set for TDK SA to achieve optimum performance".

"The wear on recording heads is significantly reduced by using TDK Super Avilyn as compared with any Chromium Dioxide tape."

"Chromium Dioxide tape is not recommended for use with any Nakamichi tape decks."



From the report by Louis A. Challis & Associates Pty Ltd. Consulting Acoustical & Vibration Engineers, NATA laboratory.

"TDK Super Avilyn Tape looks like being one of the most important advances in tape formulations in the mid-seventies."

The TDK Philosophy

To offer cassettes which are capable of capturing and faithfully reproducing the real essence of music and other complex sounds – the subtle harmonics and transient phenomena which give complex musical passages their unique "human" qualities: richness, depth, timbre and other natural sound characteristics.

There is a TDK TAPE for every occasion.

		Lengths Available
TDK DYNAMIC	Excellent quality at moderate prices	DC45, DC60, DC90, DC120 DC180
TDK SUPER DYNAMIC	"The Tape-of-the-pro" – The one that turned the cassette into a true high fidelity medium	SD C45, C60, C90, C120
TDK SUPER AVILYN	"The state of the art" performance	SA C60, SA C90
TDK EXTRA DYNAMIC	For discriminating audiophiles who demand the very best	ED C45, ED C60, ED C90

**DON'T ASK FOR TAPE – ASK FOR TDK
FOR THE SOUND YOU FEEL AS WELL AS HEAR**

TDK SA breakthrough in tape technology

Super Avilyn's performance exceeds that of Chromium Dioxide formulation which previously was the best choice for linear high frequency response and high-end S/N, but CrO2 suffered from reduced output in the middle and low frequencies (SA provides 1.5-2db more output than the best CrO2 in those ranges, equal output at high frequency).

SA also outperforms the ferric oxide tapes (regular or cobalt energized) which are unable to take full advantage of the noise reduction benefits of the CrO2 equalization because their high end saturation characteristics are not compatible with this standard (they require 1EC 120ms, normal or high EQ).

The net result of SA's characteristics and this EQ difference is a tape with an impressive 4-5db S/N gain over the latest top-ranked high output ferric oxide tapes and more than 10-12 db S/N gain over many so-called low noise ferric oxide tapes.

Ask for TDK SA Cassettes



TDK

Australian Distributor
Convoy International Pty. Ltd.
4 Dowling Street,
Woolloomooloo 2011 358 2088

NEWS DIGEST



RAY HAMILTON

TELEPHONE FOR THE DEAF

Dr. James C. Marsters (who is himself completely deaf) has developed a system which enables the deaf to 'talk' to one another. With the help of a deaf electronics engineer, Dr. Marsters has worked out a communications system that combines the telephone with a teletypewriter. By tapping the teletypewriter keys, messages are carried back and forth from one terminal to another via the telephone link. The system is called 'Phonetype'.

Special features of the system are the flashing lights which make a deaf person aware of the 'phone ringing. If the person is both blind and deaf, fans are used for this purpose and a special Braille typewriter is used. The newest feature, however, is a portable keyboard which enables a deaf person to use any phone by placing the receiver into the device and sending the message in the normal way.

Selecta Game

Somebody once described the life of a TV critic as 'looking through a glass-bottomed boat as it passes down a sewer'.

Maybe this explains the quite extraordinary worldwide sales of TV games — having spent so much on a set you might as well use it for something!

Innumerable readers have asked us to publish a TV games project — we held off doing so for some time as we felt that the complexity of such a project using discrete components was too great. We also knew that special-purpose ICs would soon be available and that these would simplify the games enormously as well as enabling more facilities to be offered.

The final result, published in this issue, is a most versatile unit featuring on-screen scoring as well as offering a wide choice of games. It can even be modified to include rifle firing practice.

Electronics Today would like to thank Dick Smith Electronics for their very considerable help and encouragement during the development of this project. (Readers will be pleased to know that Dick already has a kit available for this project. It is advertised on page 43 of this issue.)

Golf Scores by Computer

The 105th British Open Championship, held at Southport this summer, was the first time that the R and A golf club has employed computer terminals to inform spectators of scores.

The club hope to eventually have a system to provide detailed information on the progress of the championship, backed up with analyses of the play and statistics on the players.

THAT'S APPRECIATION!

When putting together our Mini-Mart page we were pleasantly surprised to find out that a March 1975 issue of ETI is currently worth \$5! Perhaps we should put up the cover price of new issues now we know what you are prepared to pay?

Bubbles coming

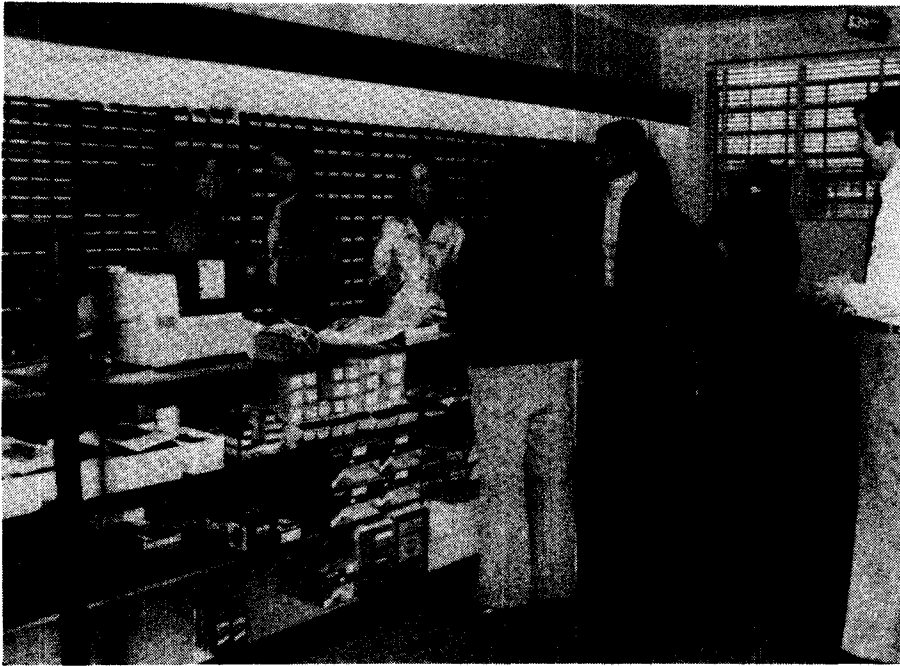
Texas Instruments will have samples of their bubble memory before the end of the year. These will be early production models — not laboratory prototypes, according to TI. The single memory chip is organised in 144 registers of 641 bits. Besides the storage material the device contains two permanent magnets to supply the constant direct-current bias needed to preserve the bubbles when the power is removed. Wire coils supply a rotating magnetic-drive field to propagate the bubbles after they've been generated within the epitaxial film. The memory in a 14-pin plastic DIP.

Watch with a message

A new digital watch from Hughes will have, in addition to the usual functions, a pre-programmed message to flash up on its nine-segment LED displays. The message is limited to a maximum of four five-letter words, which are read sequentially. There is no masking charge for orders over 1000.

Calling Brisbane Hobbyists

We draw your attention to the item in Mini-Mart calling for the formation of a general electronics hobby club in your area. If anyone is interested in using the magazine to set up a club in any other area please let us know.



ELECTRONICS ENTHUSIASTS NEW STORE

Ian Purdie's new Sydney Western suburbs electronics parts business is expanding rapidly.

An electronics enthusiast himself, Ian knows precisely what the hobbyist

needs. The company — Electronic Enthusiasts Emporium is located at Shops 2 & 3 Post Office Arcade, 7-10 Joyce St, Pendle Hill. There's plenty of parking at the rear of the premises.

EYES IN THE BACK

A Puerto Rican company has developed a system to enable the blind to see by creating images on a matrix of electrodes pressed against the skin of the back. Blind people using the system have been able to walk unaided through obstacle courses and one man can use the system to read one-inch high lettering.

The visual images are picked up by a lens on a spectacle frame and fed via a fibre-optic link to a vidicon tube housed on the main pack worn around the waist. The electrical signals are processed and fed to 3,600 electrodes on the inside of a fabric vest. The electrode voltage is nominally 6 V.

EMI DATA

The UK Company EMI Limited has announced the formation of a new division, EMI Data, as part of its Industrial Electronics operation, to further develop and market a range of high-security products based on the use of visible or invisible encoding using magnetic materials.

EMI Data uses secure magnetic materials to carry coded information to validate cards and documents used in cash and credit transactions, travel, restricted entry or identification situations.

Optical link in UK cable TV

Corning Glass Works tell us that Rediffusion Limited has been using a 1.427-kilometre length of optical cable to carry cable TV programmes to 34000 subscribers in the Hastings area (UK). The optical cable was inserted into the existing network in place of metallic-conductor cable.

In England, the optical cable for Rediffusion was installed in 75mm or 100 mm ducts in two sections, the first connected to the second at a distance from the start of 791 metres. Rediffusion engineers reported that the ducts were extensively silted and are generally rather wet. The ducts already carried up to a dozen other cables associated with the system's trunk route and its local distribution network. No special precautions were observed to pull the optical cables through, although quite considerable force was required. The installation was completed in two days without cable damage, the engineers said.

The conversion from electrical to light signals takes place in a Plessey gallium arsenide light-emitting diode. At the other end, a photodiode changes the light signals back to electrical signals.

ETI NEDERLAND!

Electronica Top Internationaal is now on sale in Dutch newsagents — the first issue in Dutch was published last month. That extends the ETI empire to four editions — in England, France, Holland and of course here where it all started in Australia.

In the next few months we expect to announce yet another edition, but we're keeping it a secret for the time being. At this rate we'll soon have the world completely covered and then we can start on the 'Electronics Today Intergalactical' mentioned last month.

BIG-DISPLAY CLOCK MODULE

A digital clock with LED display showing numerals .84-inch high is now available in quantity from National Semiconductor.

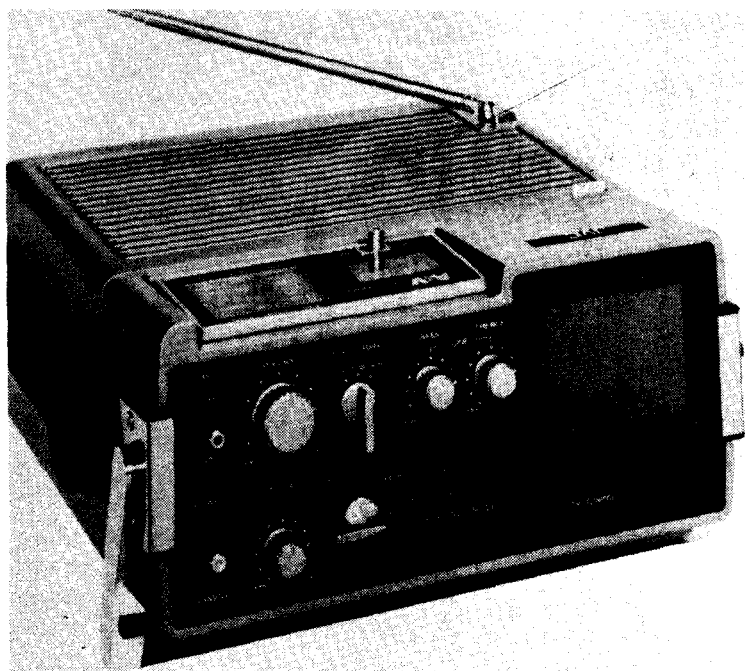
The "MA1010" series clock modules, with LS1 chip, power supply and other discrete components on a single printed circuit board, feature a four-digit .84-inch Light Emitting Diode (LED) display. According to NS this represents a larger numeric display than on any digital clock module previously manufactured.

The user only needs to add a transformer and switches to construct a pretested digital clock for application in a clock radio, digital alarm clock or instrument panel clock.

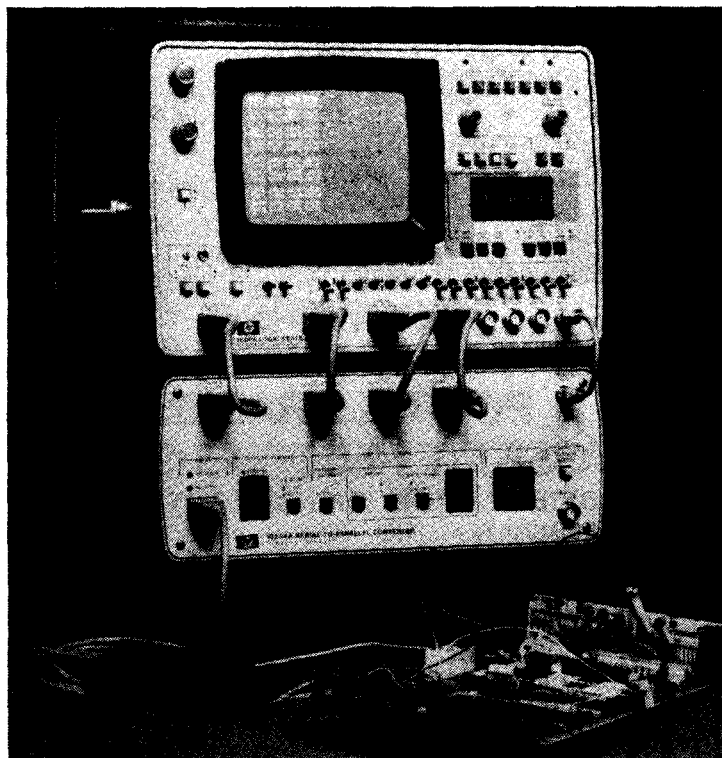
Direct, non-multiplexed drive for the LED display eliminates RF interference, which makes the module easy and economical to use in clock radios and hi-fi systems.

Features include indicators for "alarm-on" and "p.m.", a blinking colon to indicate interruption of power, "sleep" and "snooze" timers, and capability for a variable-brightness control. Time-setting is made easy for the user by providing "fast" and "slow" scanning controls.

Alarm clock options include a transistor oscillator circuit that is capable of driving an 8 ohm loudspeaker, or may be used with an inexpensive ear-phone type audio transducer.



The JVC 3050 AUL is a portable FM/MW/SW radio with a TV. It uses a 3 inch screen and a 5 inch speaker. It comes with accessories for operation from car battery or mains. The recommended retail price is \$269.

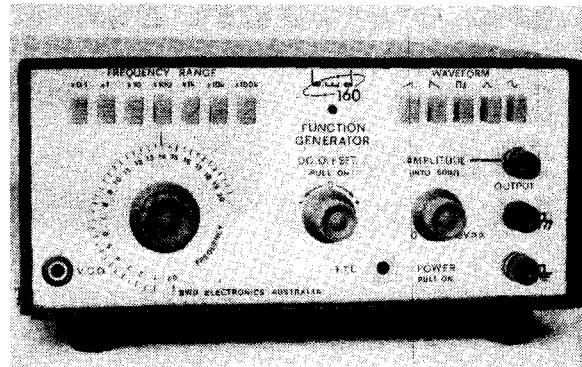


HP SERIAL-TO-PARALLEL

A new serial-to-parallel converter from Hewlett-Packard, an accessory for HP Logic State Analysers, now makes it possible to apply the analysers' full power to the serial data domain (each "word" treated serially bit by bit on a single line rather than all bits treated simultaneously on parallel lines).

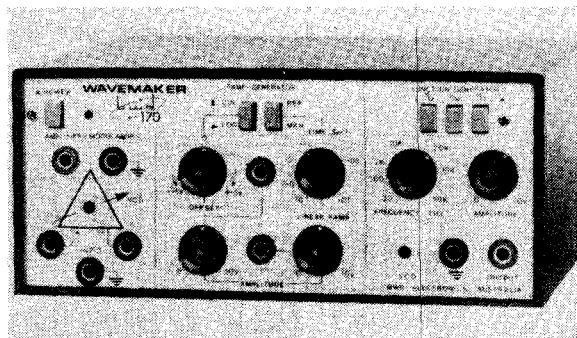
FUNCTION GENERATORS

If you are concerned with response measurement of audio waveforms two instruments recently developed by BWD Electronics will be worth inspection.



The first of these is the BWD 160A Function Generator, a moderately priced instrument providing selection of 12 different waveforms and 5 simultaneous outputs. Any 4 decades of the frequency range can be swept by an externally-applied sweep. This job is simplified by a companion instrument, the BWD 170 'Wavemaker' which supplies log or linear sweep together with facilities for amplitude or balanced modulation and frequency doubling.

The pulse and ramp outputs are set at 1/10 the indicated frequencies and are available with rising/falling ramp or positive/negative pulse. Five outputs consist of the three basic waveforms at a 1 V p-p level on the rear panel, TTL and the main variable amplitude output (10V pp into 55ohms) on the front panel.



The BWD 170 'Wavemaker' is a unique multi-purpose instrument that will generate, modulate, sweep, multiply or amplify waveforms. It supplies sine, square and triangle waveforms which may be swept over 20 Hz to 50 kHz by the ramp generator, or AM or FM modulated by an external modulation source. With a signal source such as the BWD 160A, it provides log or linear ramp to sweep the VCO and linear ramp to drive an oscilloscope or recorder. It modulates internal or external signals to 10 MHz and its frequency doubling capacity produces a variety of wave shapes.

Technical data sheets for both instruments are available from BWD Electronics Pty Ltd, Miles Street, Mulgrave Vic 3170.



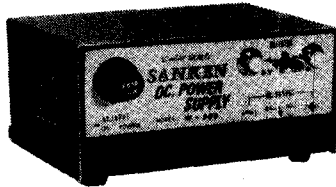
XENON
WORLD IMPORTS

Import Agents

P.O. Box 33, Warradale, South Aust. 5046.
Tel: 296-1033

DC POWER SUPPLY

**SANKEN
SOLID STATE
MODEL W-300**



REGULATED DC POWER SUPPLY

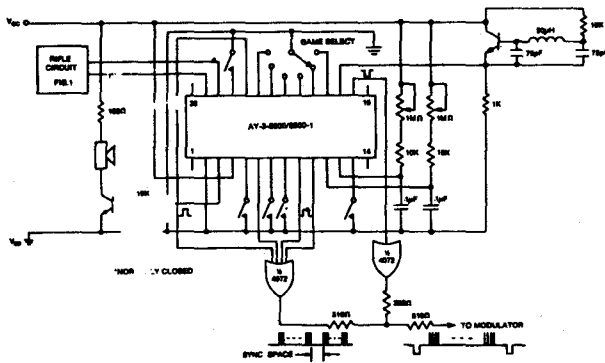
This unit is a compactly designed one with a fully transistorized circuit for use with car stereos, transceivers and other electronic equipment where a well stabilized DC power source is required as well as in the laboratory or factory. This item can also be used as a charger for car or similar batteries and as a stable power supply for testing electrical circuits and other DC items.

LIMITED OFFER ONLY \$29.00
(incl. return postage)

SPECIAL T.V. GAME I.C.

from G. I. Ltd. in U.S.A.

5 games 3 sounds in 1 Chip. New . . . but removed from circuit boards. Guaranteed working. Full circuit included in price.



Part No. AY-3-8500-1 Chip
1 to 50 \$9.00 each
50 to 100 \$7.00 each

**ONCE ONLY OFFER
TILL SOLD OUT**

SAVE

with an **ARLEC**

plug pack

TRIPLE VOLTAGE AC ADAPTOR

**MAKES
DRY
BATTERIES
OBSOLETE**

**POWERS
ELECTRONIC CALCULATORS,
PORTABLE RADIOS,
CASSETTE RECORDERS, ETC.**

DIRECT FROM MAINS ELECTRICITY

- ELIMINATES THE NEED FOR COSTLY DRY BATTERIES IN PORTABLE EQUIPMENT.
- PLUGS DIRECTLY INTO MAINS POWER SOCKET.
- SELECTOR SWITCH FOR 6, 7½ AND 9 VOLT EQUIPMENT.
- 300 mA D.C. OUTPUT.
- FITTED WITH COAXIAL CONNECTING PLUG, TO FIT MOST RADIOS, RECORDERS, ETC.
- DOUBLE INSULATED FOR ABSOLUTE SAFETY.
- APPROVED BY ELECTRICITY SUPPLY AUTHORITIES.
- 12 MONTHS GUARANTEE.

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30 Lexton Road, Box Hill, Vic., 3128, Australia. Telex 32286.

RESULT OF SEPTEMBER'S CONTEST

The Unitrex 90SC offered as a prize in our September contest goes to J E Goris of North Ryde, NSW.

The original puzzle read 'Find the smallest number which gives a remainder of eight when divided by nine, a remainder of seven when divided by eight, and so on down to giving a remainder of zero when divided by one. However at the proof-reading stage we noticed that the word 'smallest' was missing. Our immediate thought was to have the word inserted but then we thought no, we'll leave it out and see what reponse we get (and we decided to accept as valid any entry, whatever the magnitude, which satisfied the criteria).

When the entries came in interesting things started to happen. Firstly, there were many more entries than usual — in all we received over five hundred. About 50% gave a single solution and the others gave either a list of solutions or a general formula for all the solutions (of which there are an infinite number). Very few entries were incorrect (about a dozen).

The most popular general solution to the puzzle was based on $9! - 1$, or $(9 \times 8 \times 7 \times 6 \times 5 \times 4 \times 3 \times 2 \times 1) - 1$, which works out as 362879. In general this becomes $i9! - 1$, where i is any positive integer. This solution does not, however, give all possible solutions.

The smallest positive number which solves the puzzle is the lowest common denominator minus one: $(9 \times 8 \times 7 \times 5) - 1$, which gives 2519. For general solutions based on this we received $2520n - 1$, $2520i - 1$, $3 \times 2 \times 3 \times 2 \times 4 \times 3 \times 5 \times 4 \times 7 \times 5 - 1$, 2519 plus 12600n, $2520n - 1$, 2519n plus $n - 1$, where i or n can be any positive integer. Some of these are correct and the last one looks like it might include all positive solutions (unfortunately we're too busy getting the magazine out to spend time investigating the validity of these formulae).

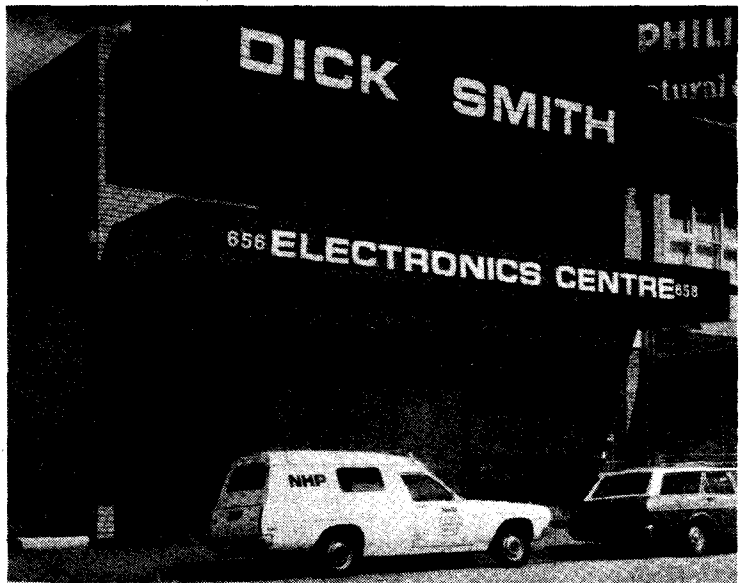
Ten entrants sent in computer printout and the programs they used to find the solution. Several of these included a few hundred suitable answers; one person sent in ten thousand two hundred!

The answer we originally were thinking of was 2519, which, as one reader pointed out, also gives a remainder of nine when divided by ten. However had we said the smallest number when we published the contest we would still have had trouble: Three readers sent in 'minus one' as the solution. This appears incorrect when you say 'nine goes into minus one zero times, zero times nine is zero, the remainder is minus one' (the same goes for all the numbers specified except "one" which goes into minus one 'minus one' times, leaving a remainder of zero).

So, it seems, it is only in the case of division by 'one' that the requirements are fulfilled. But look what happens when you say the other numbers go into minus one 'minus one' times: nine goes into minus one minus one times, nine times minus one is minus nine, the remainder is (-1) minus (-9) which gives eight! One reader actually suggested a series of negative solutions. Fortunately the randomly-selected winning entry wasn't one of these negative entries so we didn't have to decide on their validity.



Dick's new Brisbane Store.



Dick's new Melbourne Store.

DICK SMITH'S NEW STORES

Dick Smith Electronics has opened two new stores; one each in Brisbane and Melbourne. The group now has five stores down the eastern coast of Australia, as well as many dealers across the

mainland and Tasmania.

The Brisbane address is 166 Logan Road, Buranda, and the Melbourne address is 656 Bridge Road, Richmond.

AUDIO GROUP 4 YEARS OLD

In November the IREE Audio Group celebrates its 4th Anniversary.

Over 40 meetings have been held by the Group since it was first launched at the end of 1972. Membership continues to increase and interested readers can obtain a membership application form from the Audio Group Secretary, Mr. Peter Garde, at the School of Electrical Engineering, The University of New South Wales, PO Box 1, Kensington, NSW, 2033, Phone 662-2829.

China's TV

China has just put a satellite into orbit to finalise its national TV network. Now Peking can broadcast to 23 of its provinces.

Attractive Antenna

Two of the CB antennas on sale in the US by Turner use a novel mounting idea. At the base there is a big magnet, capable of a 41 kg pull. The stainless steel whip is 122 cm long.

No. 1 rated!

In US by "Consumer Report"

CORVUS

Thoroughly recommended in Australia by a major electronics publication. Electronic Concepts Pty. Ltd. is proud to introduce the exclusive Corvus 500.

With MOSTEK* single chip technology, the new Corvus 500 is the first non-Hewlett-Packard calculator with Reverse Polish Notation. 10 addressable memories, 4 level roll down stack to be introduced. If you compare the Corvus 500 feature by feature with the HP45, you will find striking similarities. There are also some important differences.

*MOSTEK is one of America's advanced LSI (Large Scale Integration) chip manufacturers.

	Corvus 500	HP 45
RPN (Reverse Polish Notation)	Yes	Yes
Memory Store and Recall. 10 Registers	Yes	Yes
4 Level Stack, Rotate Stack	Yes	Yes
10 MEMORY EXCHANGE WITH X	Yes	No
Log. LN	Yes	Yes
Trig (Sine, Cosine, Tangent, INV)	Yes	Yes
HYPERBOLIC (SINH, COSINH, TANH, INV)	Yes	No
HYPERBOLIC RECTANGULAR →	Yes	No
y ^x , e ^x , 10 ^x , √, 1/x, x!, x←y, π, CHS	Yes	Yes
↔ through INVERSE	Yes	No
GRADIANS	No	Yes
DEGREE-RADIAN CONVERSION	Yes	No
Degree Radian Mode Selection	Yes	Yes
DEC-DEG-MIN-SEC	No	Yes
Polar to Rectangular Conversion	Yes	Yes
Recall Last X	Yes	Yes
Scientific Notation, Fixed and Floating	Yes	Yes
Fixed Decimal Point Option (0-9)	Yes	Yes
DIGIT ACCURACY	12	10
DISPLAY OF DIGITS	12	10
%, √, %	Yes	Yes
GROSS PROFIT MARGIN %	Yes	No
Mean and Standard Deviation	Yes	Yes
Σ +, Σ -	Yes	Yes
Product - Memories	Yes	Yes
C.F. DIRECT CONVERSION	Yes	No
F.C. DIRECT CONVERSION	Yes	No
LIT.GAL. DIRECT CONVERSION	Yes	No
KG.LB. DIRECT CONVERSION	Yes	No
GAL.LIT. DIRECT CONVERSION	Yes	No
LB.KG. DIRECT CONVERSION	Yes	No
CM.INCH DIRECT CONVERSION	Yes	No
INCH.CM DIRECT CONVERSION	Yes	No

As you can see, the Corvus 500 is a lot more calculator for \$95.

Price \$95.00

Mail charge \$2.50

Sales Tax exempt \$85.00

For sales tax exempt purchases, please supply number or certificate.

We have listed some of the many features, but let's amplify on some highlights:

1. RPN (Reverse Polish Notation) "COMPUTER LOGIC" and 4 LEVEL STACK.



ELECTRONIC CONCEPTS PTY LTD

Ground floor, Cambridge House,
52-58 Clarence St., Sydney NSW 2000.
(02) 29 3753-4-5



Your problem is solved the way it is written, left to right sequence, eliminating restructuring, unnecessary keystrokes, and the handicap of having to write down intermediate solutions. And all information is at your disposal — just roll the stack (R) to any intermediate information desired. You arrive at your solution faster, more simply and, therefore, more accurately.

Perhaps at this point we should address ourselves to the controversy between algebraic entry and RPN. One question we must ask is why proponents of algebraic entry always use an example of sum of products and never an example of product of sums:

$(2 + 3) \times (4 + 5) =$
 $\text{Algebraic } 2 + 3 = \text{MS } 5 + 4 = \text{XMR} =$
TOTAL 12 keystrokes (SR51, add 2 more keystrokes)
RPN: 2 Enter 3 + 4 Enter 5 + x
TOTAL 9 keystrokes

2. THE CORVUS 500 and HP-45 HAVE 10 ADDRESSABLE MEMORY REGISTERS, 4 LEVEL OPERATIONAL STACK, and a "LAST X" REGISTER (10th Mem. Reg.). With 10 addressable memories, you have access to more entries, or intermediate solutions; less remembering, or writing down. YOU have to do. And less chance for error. The stack design also permits X and Y register exchange, and roll-down to any entry to the display for review or other operation. The "last x" register permits error correction or multiple operations when a function is performed, the last input argument of the calculation is automatically stored in the "last

x" register, which can be quickly recalled to correct an error, or to perform another operation using the same number.

3. DIRECT HYPERBOLIC and HYPERBOLIC RECTANGULAR to POLAR, and INVERSE. For those of you electronic and computer science engineers who require access to this specialised application, the Corvus 500 solves "your" problems.

4. A WORD ABOUT CORVUS 500 12-DIGIT DISPLAY AND ACCURACY. Finally you have displayed 12 digit accuracy in business format and 10 + 2 in scientific notation. LED is manufactured by Hewlett Packard.

FOR THE FIRST TIME you can raise the number 10 to 199th power or calculate Factorial (x!) of up to 120. Unbelievable!

5. DIRECT FROM AND TO METRIC CONVERSION SAVES VALUABLE KEYSTROKES.

WHAT ABOUT CONSTRUCTION? With so many features, the next most obvious question must be in regard to the quality of the unit itself. We are proud to report the Corvus 500 to be double injected moulded, with "tactile" feedback keyboard. The compact, contoured case is 5 1/2" long by 3" wide by 1 1/4" high and weighs just 8 oz. The COMPLETE CORVUS 500 for \$95.00 includes:

- Rechargeable and replaceable Nickel Cadmium batteries. Optional 3AA batteries.
- Adaptor/Charger.
- Owner's Handbook.
- Soft carrying case.

The Corvus 500 is warranted by the manufacturer against defects in materials and workmanship for one year from date of delivery.

For those of you who have the HP-21 or 45 or any other advanced calculator on order, aren't you glad you still have the opportunity to take advantage for the release of the Corvus 500 for \$95.00? Hurry! Order yours today.

AN INVITATION:

Electronic Concepts is proud to offer this exciting Corvus 500 as well as other Mostek based calculators and digital watches as exclusive importer of Corvus Brand products for Australia.

You, our discerning reader will no doubt recognise the tremendous price/performance value on offer. By mailing the order coupon today we can assure you of early delivery — and should you not be satisfied, you may return the unit to us with full money back guarantee within seven (7) days. Or better, convince yourself of the real quality and value of our Corvus range, just visit our conveniently located showroom in Cambridge House, Clarence Street, just behind Wynyard exit (York Street), or phone 02-29-3755 for more information.

Other Corvus models on offer:

Corvus 600 Financial Genius \$80.00

Corvus 615 Business

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Debney's Park is a complex in the suburb of Flemington, on the fringe of the city, and contains a number of high-rise blocks. The closed circuit TV is seen as a means of combating the loneliness which sometimes afflicts people living in high rise communities.

'ELECTRONICS IN MEDICINE' LECTURES

"Electronics in Medicine" is the theme of a series of public lectures in the five mainland State capitals and Canberra and Summer Schools in Sydney and Melbourne between November 29 and December 22.

"Medical Electronics — A Modern Coalition Against Ancient Constraints" is the title of the public lectures by Prof. Sayers at the University of New South Wales on December 2; University of Brisbane, De-

ember 8; University of Melbourne, December 14; University of Adelaide, December 20; University of Western Australia, December 21 and the Academy of Science, Canberra, on December 3.

In Sydney, a full 4 day Summer School Course for electronics engineers and students specialising in medical electronics will be held at International House, University of New South Wales, on November 29, 30, December 1 and 2, and a half day 4 afternoons Summer School Course for members of the medical profession and medical research workers and students on December 6, 7, 9 and 10.

In Melbourne, a four-day (full day) Summer School will be held for engineers and medical workers in both disciplines at the University of Melbourne on December 13, 14, 16 and 17.

Further information and registration forms may be obtained from the General Secretary, IREE, 157 Gloucester Street, Sydney, 2000. (Telephone 27-1039).

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Subject matter for the course will include digital electronics, small computer architecture, microprogramming, interface design, data transmission and error control and advanced control theory.

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For further details please contact Mr W A Evans or Mr E K Walker, Footscray Institute of Technology, Department of Electrical & Electronic Engineering, Ballarat Road, Footscray, Vic 3011 (Telephone 689-3400).

ERRATA

SIMPLE AMPLIFIER. Project 061. October 1976, page 64.

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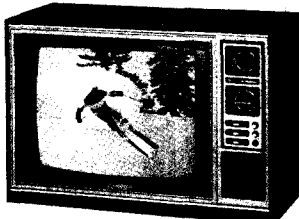
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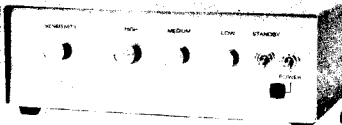
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Film resistors

Two previous articles in our components series have dealt with resistors. This article now looks at carbon-film resistors and the next part will cover metal-film types.

FILM RESISTORS ARE MANUFACTURED by forming a deposit of an appropriate resistive material, usually carbon, carbon-boron or some metallic oxide, on a ceramic former, usually a tube or rod. A helical groove is then cut in the film coating. The groove forms the resistive coating into a long continuous path resulting in a compact resistor that can have a value up to 100 megohms. Terminations are made in a variety of ways. Metal end caps may be forced over the ends of the ceramic rod, contacting the deposited film. Leads are attached to the caps by soldering or spot-welding. In some types, the ends of the coated ceramic rod are metallized and leads are wrapped around the metallized portions and soldered. The component is then coated in a suitable lacquer for protection.

Typical construction of a film resistor is illustrated in Figure 1.

Thick-film resistors are a special type of film resistor. They are generally con-

structed by depositing the resistive material on a ceramic or aluminium-oxide substrate. A portion of the film coating is then removed, according to a predetermined pattern, to provide a long resistive path between the resistor terminals. Typical construction of one style of thick-film resistor is illustrated in Figure 2. This style is obtainable as a 'fusible' resistor. When overloaded, the substrate cracks, ensuring an open circuit which reduces the possibility of further circuit damage, physical or electronic. These thick-film resistors occupy a minimum of space on a printed circuit board and can dissipate considerable power owing to their large surface area and high hot-spot temperature (150°C).

Thick film resistors are also made in appropriate groupings on a small substrate and encapsulated in a standard DIL IC package. Certain values of resistance are standard in digital circuitry and this style is used in such

applications (for example, as the 'weighting' resistor network in a digital-to-analogue converter). Another application is for 'pull-up' resistors for open-collector logic gates.

Thin film resistors are constructed in a similar fashion but on a considerably smaller scale. They are primarily used in IC manufacture. Some thin film resistor networks are available in standard DIL integrated circuit packages and these find application in digital circuitry.

There are four basic types of film resistor:—

- (a) Carbon Film
- (b) Metal Film
- (c) Metal Oxide Film
- (d) Metal Glaze (Cermet)

Carbon Film Resistors

These resistors are manufactured by a 'cracking' or pyrolytic process where a hydrocarbon vapour at high temperature is decomposed onto a special

TABLE 1. General Characteristics of Carbon Composition Resistors

Rated Wattage @ 40°C	Max. Working Voltage	Max. Operating (= hot-spot) Temperature		Critical Resistance		Uninsulated Types		Insulated Types		Typical Resistance Range		
		Commercial	Mil.	Commercial	Mil.	Length	Diameter	Length	Diameter			
0.5	0.125	500	150	107	130	2 M	180 k	6.4 mm	2.3 mm	3.7 mm	1.6 mm	4.7 Ω - 1 M
0.75	0.25	700	250	107	130	3M9	250 k	10.8 mm	4.1 mm	9.5 mm	2.4 mm	2.2 Ω - 10 M
1	0.35	1000	—	107	130	2M7	—	10 mm	3.5 mm	9.5 mm	4.8 mm	10 Ω - 4M7
2	0.5	1000	350	107	130	1M8	250 k	18 mm	6 mm	11.7 mm	5.8 mm	2.2 Ω - 100 M
2	1	1000	500	107	130	1 M	250 k	29.8 mm	7.7 mm	19.1 mm	6.4 mm	10 Ω - 1 M
2	2	1000	500	107	130	470 k	120 k	29.8 mm	7.7 mm	35 mm	8 mm	10 Ω - 1 M

- (1) Rated Wattage assumes voltage limit not exceeded.
- (2) Max. Working Voltage assumes wattage rating not exceeded.
- (3) Max. Operating Temperature is that due to ambient temperature plus temp. rise due to power dissipation. No power can be dissipated by a resistor if the ambient temperature equals the hot-spot temperature. The hot-spot temperature for commercially-rated carbon composition resistors is usually between 105 and 110°C.
- (4) Sizes given are body sizes for axial-lead types.

Film resistors

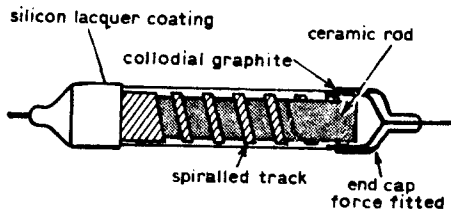
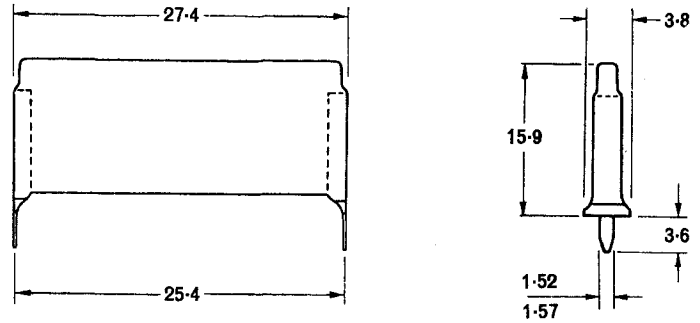


Fig. 1. Typical construction of a film resistor.



Dimensions in mm

Fig. 2. Example of fusible-type of thick-film resistor.

ceramic rod, producing a thin carbon film on the surface. These are sometimes referred to as 'deposited-carbon' film resistors. Some types use a boron-carbon film; a boron containing gas is introduced during the cracking process. This results in a resistor that has a superior temperature coefficient over a limited range of values than the plain carbon film type.

Terminations may consist of metal end-caps forced over the ends of the element, and then axial or radial leads are attached. Some manufacturers metallize the ends of the element and solder leads to them. Sometimes a combination of the two techniques is used to improve reliability.

Protection for the element is provided in a number of ways. Numerous layers of varnish may be applied followed by a final paint coating. Some modern types are completely sealed in a silicone resin base which is impervious to moisture as well as providing excel-

lent mechanical and thermal protection. Other types may be encased in a plastic moulding or sealed in a ceramic or glass tube. The varnished types afford the least protection against mechanical damage (through handling etc) and moisture.

The voltage coefficient of carbon film resistors is very much less than that of carbon composition types, being usually less than 100 ppm/V and this rarely needs to be considered.

Carbon film resistors exhibit temperature characteristics which are superior to composition resistors, but not as good as metal film or wirewound types. Nevertheless, the temperature coefficient of carbon film resistors is quite acceptable for a wide variety of applications. Only those applications requiring a very good temperature characteristic warrant the use of the other, usually more expensive, film resistor.

As mentioned just previously, the temperature coefficient of boron-carbon

film resistors is somewhat better than the deposited-carbon types. The latter may have a temperature coefficient between +350 and -550 ppm/°C for values under 100k, and between +350 and -800 ppm/°C for values under 100k. Generally though, the TC will be negative. The variation of TC with resistance value and the sort of 'spread' that can be expected for a particular batch of components is illustrated in Figure 3 for deposited carbon resistors. The temperature coefficient of boron-carbon resistors is typically between +100 and -200 ppm/°C for values under 100k, and between -50 and -400 ppm/°C for values over 100k. The variation of TC with resistance value and the spread that might be expected is illustrated in Figure 4.

The TC of carbon film resistors is also dependant on the wattage rating due to the thickness of the carbon film used in its construction.

All resistors change their value per-

TABLE 2. General Characteristics of Carbon Film Resistors

Rated Wattage @ 70°C	Max. Working Voltage	Max. Operating Temp.	Critical Resistance	Typical Sizes		Typical Resistance Ranges	
				Length	Diameter	Deposited Carbon	Boron-Carbon
0.125	250 V	130/165°C	250 k	7 mm	2.3 mm	10 Ω - 1 M	50 Ω - 100 k
0.250	300 V	130/165°C	360 k	10 mm	2.3 mm	10 Ω - 2 M	20 Ω - 100 k
0.33 (0.5 @ 40°)	300 V	125°C	360 k	9 mm	3 mm	2.2 Ω - 5M1	
0.5	350 V	130°C	250 k	12 mm	4 mm	4.7 Ω - 5M1	10 Ω - 100 k
0.5	350 V	165°C	250 k	15 mm	4 mm	10 Ω - 7M5	
0.75	350 V	165°C	160 k	14 mm	6 mm	10 Ω - 7M5	
1.0	500 V	130°C	250 k	14 mm	4.8 mm	2.2 Ω - 10 M	20 Ω - 240 k
1.0	500 V	165°C	250 k	24.6 mm	7.2 mm	10 Ω - 15 M	
1.25	600 V	165°C	270 k	22 mm	9 mm	10 Ω - 15 M	
2.0	750 V	130°C	270 k	55 mm	7.5 mm	10 Ω - 20 M	30 Ω - 1 M
2.0	750 V	165°C	270 k	32 mm	9 mm	10 Ω - 15 M	

- (1) Rated Wattage assumes voltage limit not exceeded.
- (2) Max. Working Voltage assumes wattage rating not exceeded.
- (3) Max. Operating Temperature is equal to hot-spot temperature.
- (4) Sizes given are body sizes for axial-lead types.
- (5) Coated types and silicone resin coated types only considered.

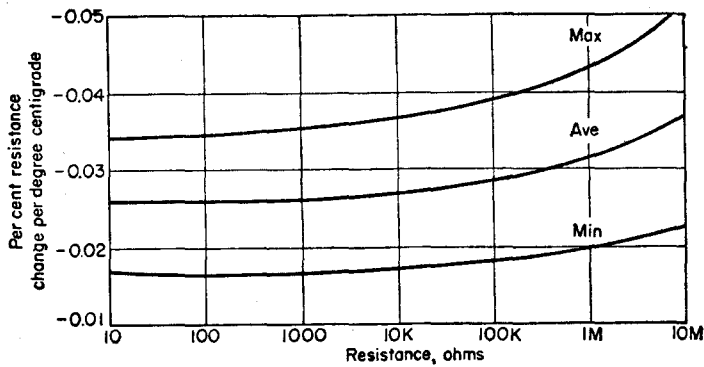


Fig. 3. Typical temperature-coefficient spread for deposited-carbon resistors.

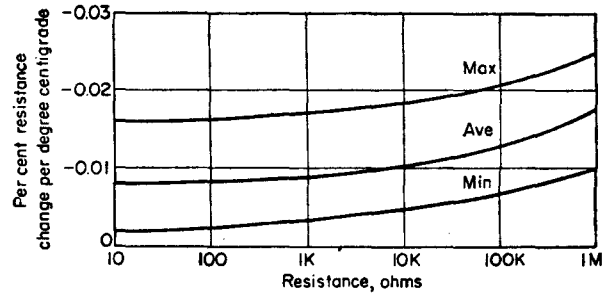


Fig. 4. Typical temperature-coefficient spread for boron-carbon resistors.

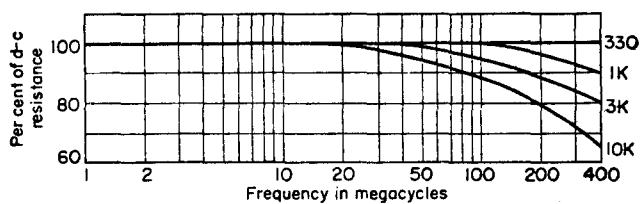


Fig. 5. Approximate frequency characteristics for 1/2-watt deposited-carbon resistor.

manently with age and use. Carbon composition resistors are the worst in this regard and may be expected to change as much as 20%. Film and wire-wound resistors are considerably better. Carbon film resistors have a stability of better than 1% which is usually more than adequate for all but the most stringent applications.

The high frequency characteristic of carbon film resistors is one of its advantages. Coated types are somewhat better than equivalent moulded or encased units. Generally speaking, the apparent value of the resistor decreases

at high frequencies. Values below 1k will maintain their resistive value well beyond 500 MHz. Even relatively high values will not show a decrease of more than 10% until well into the VHF region. This is illustrated for typical coated 1/2W deposited-carbon film resistors in Figure 5.

The noise generated by carbon film resistors is a function of the applied voltage, the thickness of the film and the length of the spiral track. Consequently, the lower value, higher wattage units generate the least noise. For values below 10k it is typically

between .08 and .5 $\mu\text{V}/\text{V}$, and for values between 10k and 100k it may be as low as 0.2 $\mu\text{V}/\text{V}$ and up to 1.0 $\mu\text{V}/\text{V}$. For values above 100k, the noise ranges from 0.5 $\mu\text{V}/\text{V}$ to 1.5 $\mu\text{V}/\text{V}$.

There are several power derating curves for carbon film resistors, dependant on size and construction. Miniature coated types have a hot-spot temperature of 120–125°C and are derated from 40°C to half their wattage rating, at 70°C, then derated to zero dissipation at the hot-spot temperature. This results in a 'dogs-leg' derating graph as shown in Figure 6. This mainly applies to the miniature 0.25W and 0.33W types which have body dimensions typically 6-7mm long and about 2.5mm diameter. Moulded style units are usually derated from 70°C and have a hot-spot temperature of 130°C, according to the derating curve shown in Figure 7. Some types have a much higher hot-spot temperature, being constructed on a special ceramic rod

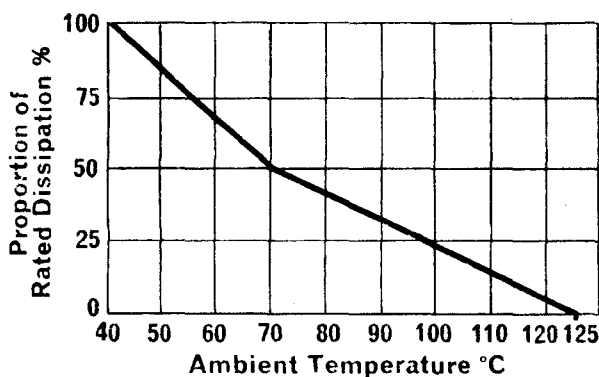


Fig. 6. Derating curve for miniature moulded carbon-film resistors.

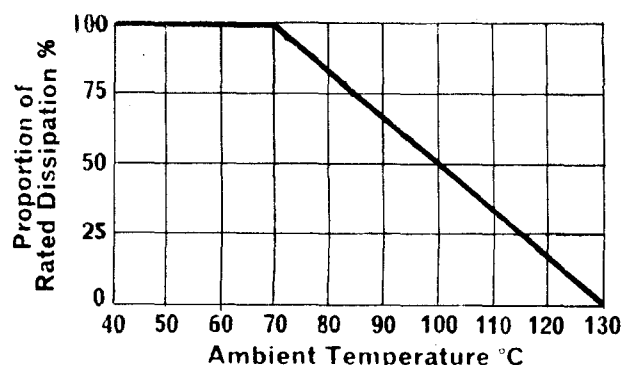


Fig. 7. Derating curve for coated carbon-film resistors.

Film resistors

and coated with a silicone resin compound which have superior heat dissipating properties. These types have a hot-spot temperature of around 165°C and are derated from 70°C, as illustrated in Figure 8. It is best to check the manufacturer's literature if the power derating characteristics are needed. Special 'carbon-alloy' types have a hot-spot temperature of 200°C but are not commonly used.

Carbon film resistors are available in ratings from 0.1W to 2W and in values that range from 10 ohms to 15M for commonly available units and up to 100M on special order. They are manufactured to tolerances of ± 0.5% (E192 series), ± 1% (E96 series), ± 2% (E48 series) and ± 5% (E 24 series).

Carbon film resistors will withstand a short-term overload of twice to 2.5 times the rated maximum working voltage. Failure is more common in the high value resistors. Irregularities in the spiral track and extremely thin film contribute to the failure of the component. The resistor may burst into

flame when it fails due to a prolonged overload.

The excellent stability and low cost of carbon film resistors, along with other desirable features such as low noise, small TC and good high

frequency characteristics have contributed to their increasing use in a wide range of electronic applications. The general characteristics of carbon film resistors is given in Table 2.

to be continued . . .

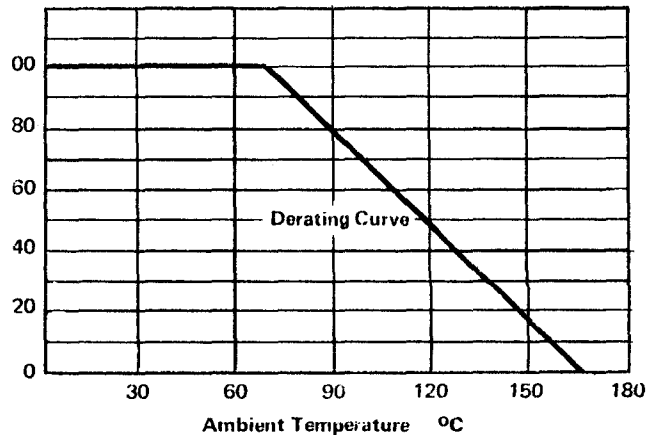


Fig. 8. Power derating curve for specially constructed carbon film resistors.

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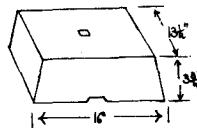
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25	25	Axial	9c
47	16	Axial	9c
47	25	Axial	13c
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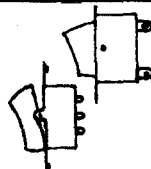
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LM307H	T05	Op Amp	.99	LM387AN	8DIL	Selected Lower Noise		LM2917N	14DIL	2907 + Zener for Dwell	2.85
LM307N	8DIL	Op Amp	.92			LM387N	2.70	LM3028AH	T05	Differential RF/IF Amp	1.40
LM308H	T05	Op Amp	3.23	LM388N	14DIL	1.5W Mono Audio Amp	1.55	LM3028BH	T05	Differential RF/IF Amp	1.90
LM308N	8DIL	Op Amp	1.70	LM389N	18DIL	LM386 + 3 transistors	1.55	LM3046N	14DIL	Transistor Array	1.15
LM311H	T05	Voltage Comparator	2.90	LM395T	T0220	IC Power Transistor	2.85	LM3064N	14DIL	TV Audio Fine Tuning	1.60
LM311N	8DIL	Voltage Comparator	2.45	LM555CN	8DIL	Timer	.95	LM3086N	14DIL	Transistor Array	.80
LM318H	T05	High Speed Op Amp	4.40	LM556CN	14DIL	Dual 555 Timer	1.70	LM3089N	16DIL	FM Receiver IF System	4.00
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LM348N	14DIL	Quad 741 Op Amp	3.11	LM709CN	14DIL	Op Amp	.99	LM4250CH	T05	Programmable Op Amp	5.60
LM349N	14DIL	Quad 741 Op Amp High Slew	3.50	LM709CH	T05	Op Amp	.99	LF352N	FET	Instrumentation Amp	NEW P.O.A.
LM358N	8DIL	Dual 741 Op Amp	1.45	LM710CN	14DIL	Voltage Comparator	1.00	LF355H	T05	FET Input Op Amp	4.00
LM370H	T05	AGC/Squelch Amp	4.00	LM710CH	T05	Voltage Comparator	1.20	LF356H	T05	FET Input Op Amp	4.00
LM371H	T05	RF/IF Amp	3.50	LM723CH	T05	Voltage Regulator	1.50	LF357H	T05	FET Input Op Amp	6.00
LM372H	T05	AM/IF Strip	4.14	LM723CN	14DIL	Voltage Regulator	1.10	LF398H	T05	FET Sample and hold	6.00
LM373H	T05	AM/FM/SSB IF Video Amp/Detector	5.06	LM725CN	8DIL	Instrumentation Op Amp	3.96	LH0042CH	T05	FET Op Amp	6.60
LM375N	8DIL	Oscillator/Buffer TTL Output	4.80	LM733CN	14DIL	Differential Video Amp	2.30	LH0070-1H	T05	10V Precision Voltage reference	5.10
LM377N	14DIL	2 Watt Stereo Audio Amp	2.75	LM741CN	8DIL	Op Amp	.75	LH0070-2H	T05	10V Precision Voltage reference	10.05
LM378N	14DIL	4 Watt Stereo Audio Amp	3.60	LM741CH	T05	Op Amp	1.14	LX5700H	T046	Temperature Transducer	5.60
LM379S	Special	6 Watt Stereo Audio Amp	6.05	LM747CN	14DIL	Dual Op Amp	1.94	MC1330P	8DIL	Video Detector	2.35
LM380N	14DIL	2 Watt Mono Audio Amp	1.60	LM748CN	8DIL	Op Amp	.93	SAK140	14DIL	Tachometer IC	2.75
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LM382N	14DIL	Stereo Preamp with R1AA/NAB	2.00	LM1351N	14DIL	FM Detector, Limiter, Audio Amp	1.90	UAA180	18DIL	12LED "Line" Driver	2.50
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				LM1458N	8DIL	Dual Op Amp	1.97				
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CMOS-a practical guide

Inherently rugged, CMOS logic has many advantages over other logic families — high noise immunity and uncritical power requirements are but two. This, the fourth article in this series, deals with multiplexing.

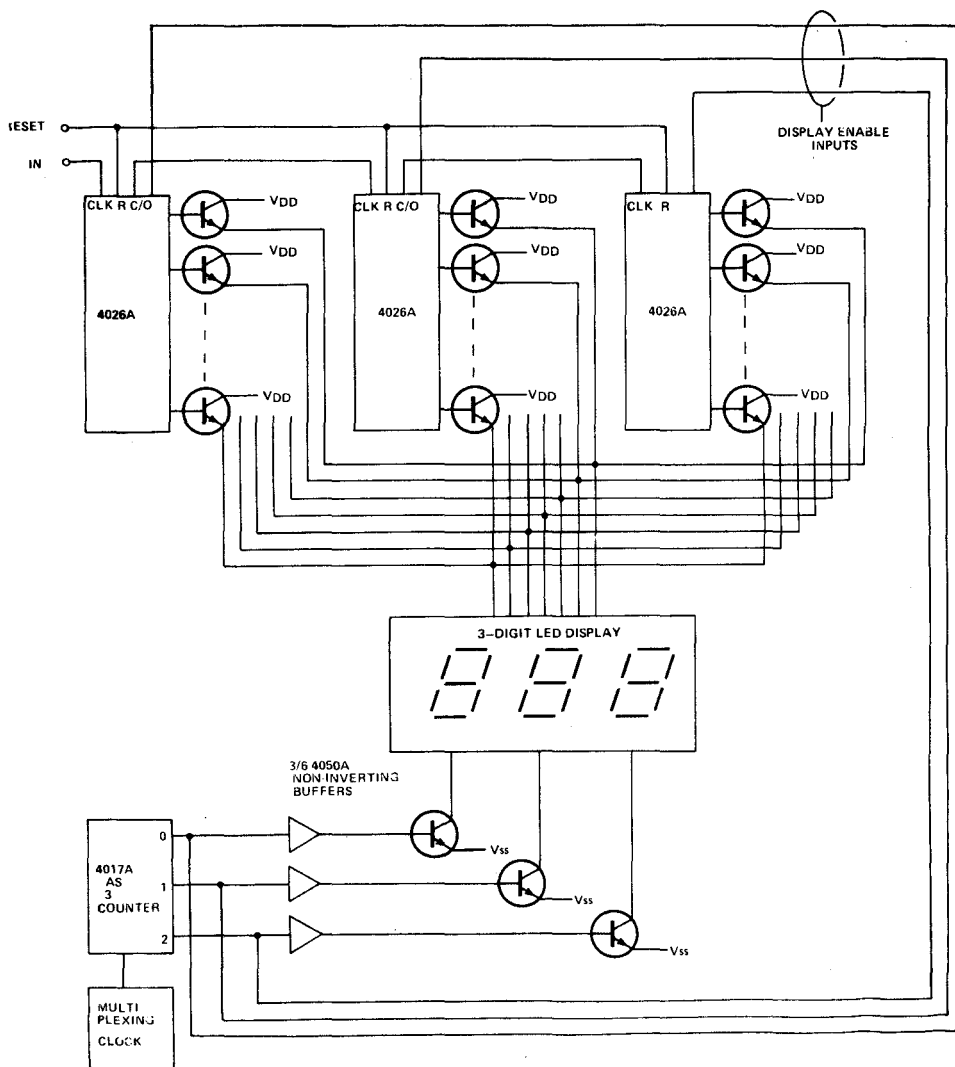


Fig. 1. A three decade counter for a 3-digit multiplexed display. Extra buffering of the digit lines may be necessary for some displays.

LIFE IS NEVER AS SIMPLE AS WE might want and there are two reasons for complicating the circuitry by using digit multiplexing (i.e. each digit is displayed for a fixed period, usually between about 10 and 30% of the time). These are that to do so is more efficient in terms of power consumption, and secondly that most multi-digit displays reduce the number of lead-outs (by giving just one set of seven segment drive lines for the complete display and one digit drive line for each digit).

This is the reason why the 4026A has a display enable input which, although the counter continues to function, cuts off the display when it is held low. The display enable output gives a replica of the input and may be used to enable other counters which are to be "on" during the same period. It also explains the presence of the "ungated C-segment" output which is used for producing some divide by "N" configurations which operate when the display is disabled.

The basic arrangement of a three decade counter is shown in Fig.1 and attention is drawn to the note that additional buffering may be necessary on the digit lines. It is also worth noting the use of a 4017 divide by three counter (using the flip-flop reset mode) to control the display.

Other types of displays which are often used are higher current LEDs such as the MAN-1 which is, in contrast to the MAN-3 a common anode device. This means it must be driven by inverting buffers as shown in Fig. 2a We have been relying here on the output current limit of the CMOS chip to limit

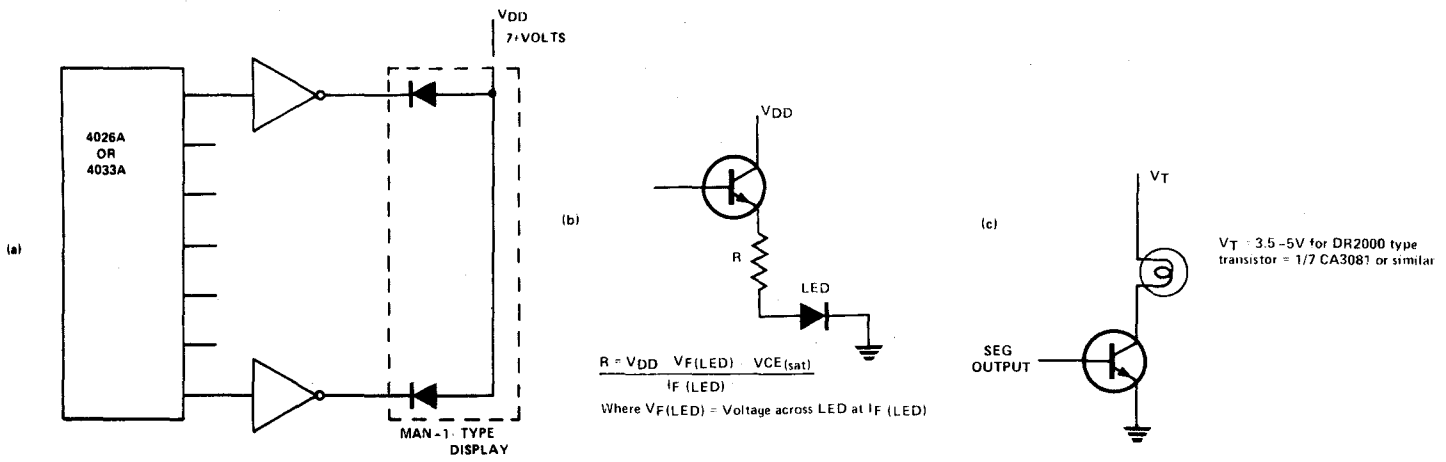


Fig. 2. Driving other displays: (a) MAN-1 type (b) example of calculation of limiting resistor (c) Numitron type incandescent display.

the forward current in the LEDs. It may be necessary to add current limiting resistors in the segment lines, particularly when transistor drivers are employed. The calculation of the value is simple given the required segment current and voltage drops (see Fig. 2(b)). In multiplexed displays the limiting resistors should, of course, be put in the common segment lines. A considerable saving in resistors in non-multiplexed displays may be achieved by putting a single resistor in the common line to each digit. The disadvantage is that the display brightness varies with the digit. Fig. 2(c) shows the technique for interfacing with "Numitron" and similar displays.

The ripple blanking facility is for

blanking leading and trailing zeroes in the display and it works as follows. Take the ripple blanking input (RBI) of the most significant 4033A on the integer side of the display low. Then take the ripple blanking output (RBO) of the IC and connect it to the RBI of the next counter and so on until the position of the assumed decimal point is reached. Follow exactly the same procedure from the least significant counter in the fractional part of the display backwards to the decimal point (see Fig. 3(a)). Of course, if the assumed decimal point is at one end of the display then half the procedure would be unnecessary. If non-significant zeroes in the places either side of the decimal point are to be displayed (so that 7 and .6 appear as 7.0 and 0.6)

then the RBIs of the two counters concerned should be taken to Vdd. Finally on these two ICs, the lamp-test facility on the 4033A just forces all segment outputs high when it is taken high.

THE 4029A and 4081A

We shall conclude our discussion of counters by looking briefly at two more devices. The 4029A is a general purpose counter which, at the price that a 7490 was a year or two ago, has most of the features of the more exotic TTL devices. Briefly, the device is positive edge triggered and advances when the clock and preset enables are both low. Furthermore it counts in binary when the binary/decade input is high and BCD otherwise. A high signal at the up/down input persuades it to count up and a low input forces it to count down. As though this were not enough, when the preset enable input is high, the Q counter outputs are forced to follow the J ("Jam") inputs. The prefix "4" in both cases indicates the most significant digit. The pinout diagram is given in Fig.

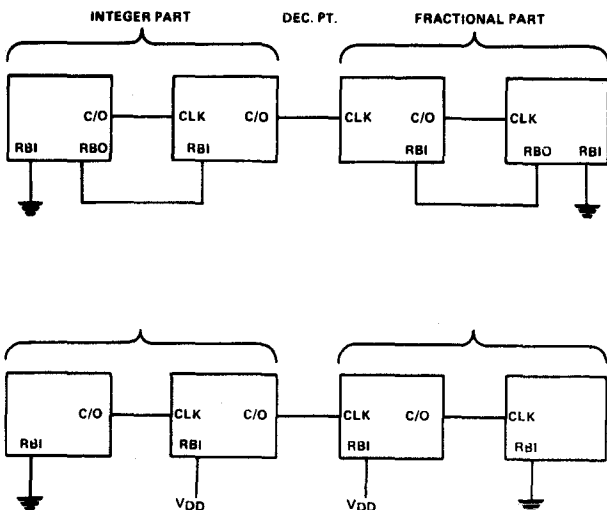


Fig. 3. Four digit counters using the 4033A with non-significant zero suppression (a) in all positions (b) in first and last position only.

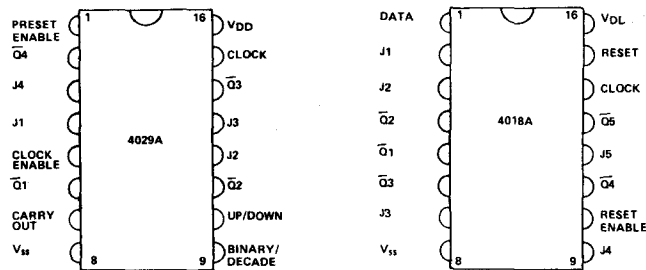
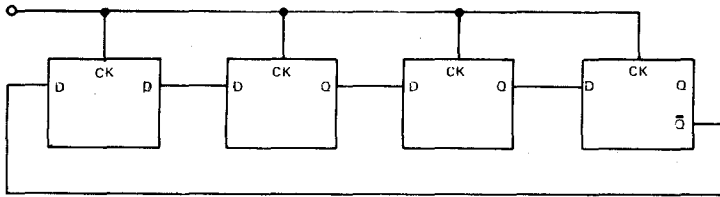


Fig. 4. Pinouts of the 4029A and 4018A.

CMOS-a practical guide



Q ₁	Q ₂	Q ₃	Q ₄
0	0	0	0
1	0	0	0
1	1	0	0
1	1	1	0
1	1	1	1
0	1	1	1
0	0	1	1
0	0	0	1

Fig. 5. Circuit diagram and counting sequence for a four stage Johnson or "twisted ring" counter.

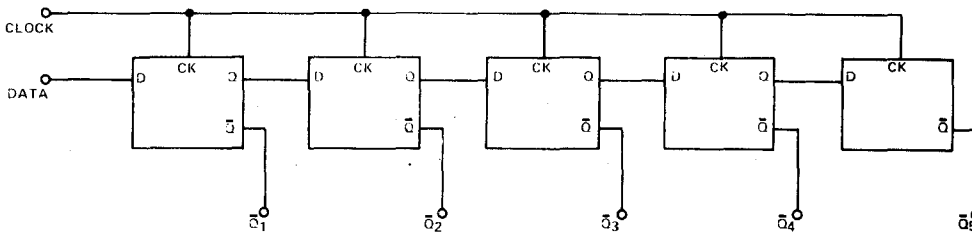


Fig. 6. Simplified internal diagram of the 4018A.

TO DIVIDE BY	CONNECT
3	\bar{Q}_1 \bar{Q}_2 DATA
4	\bar{Q}_2 _____ DATA
5	\bar{Q}_2 \bar{Q}_3 DATA
6	\bar{Q}_3 _____ DATA
7	\bar{Q}_3 \bar{Q}_4 DATA
8	\bar{Q}_4 _____ DATA
9	\bar{Q}_4 \bar{Q}_5 DATA
10	\bar{Q}_5 _____ DATA

Fig. 7. Connection of the 4018A as a divide by "N" counter. Input to clock, output waveform from DATA input is symmetric when N is even, almost so when N is odd.

4 along with that for the 4018A pre-settable divide by N counter.

There are two basic ways of producing counters. Firstly there is the chain of flip-flops each of which halves the frequency produced by the one before it. This was the principle behind the binary counters, which we considered at the beginning of this month's discussion, and also of the 4029A.

The second method is known as a Johnson counter and it is basically a shift register consisting of a chain of flip-flops with the Q output of the last counter connected back to the data input. A little patience and a pencil and paper will soon show that such a counter will divide the input frequency by 2N where N is the number of stages.

The counting sequence for a four stage counter is shown in Fig. 5 and the reader will notice that if the counter starts with contents not in the counting sequence (e.g. 1010) then the contents are always nonstandard thereafter. Thus some special gating is required. The simplified internal diagram of the 4018A in Fig. 6 is not complete. Also the Jam inputs and preset enable (which work in the same way as in the 4029A) together with the reset (which zeros all stages (Q₁ - Q₅ = 1) have been omitted for clarity.

Figure 7 shows the way to connect the 4018A to divide by all numbers from three to ten. Just as an example of how versatile this device is, one application will be considered in a totally different field from counting. By disregarding the clock, the Jam inputs and inverted data outputs (\bar{Q}) can be used as a five data latch for temporary storage, the outputs being updated to the inputs while the present enable is high.

To be continued

FERGUSON

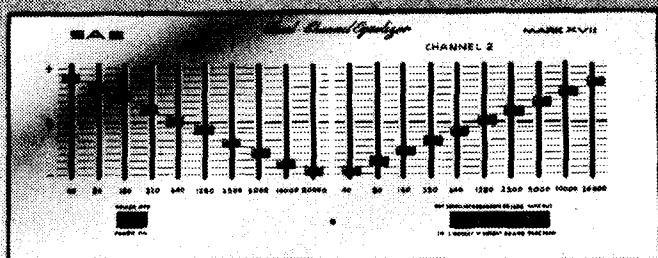
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Your tone controls are just not designed to compensate for

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We built the Mark XVII Equalizer to solve these problems and more.

- These are some of the ways:
- Individual Octave Control for each channel.
 - Long throw, oil-damped linear slide pots for greater accuracy.
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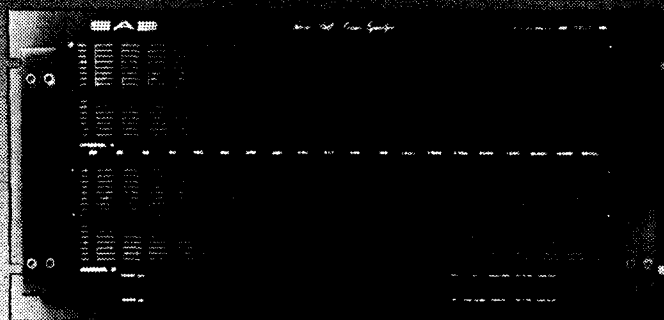
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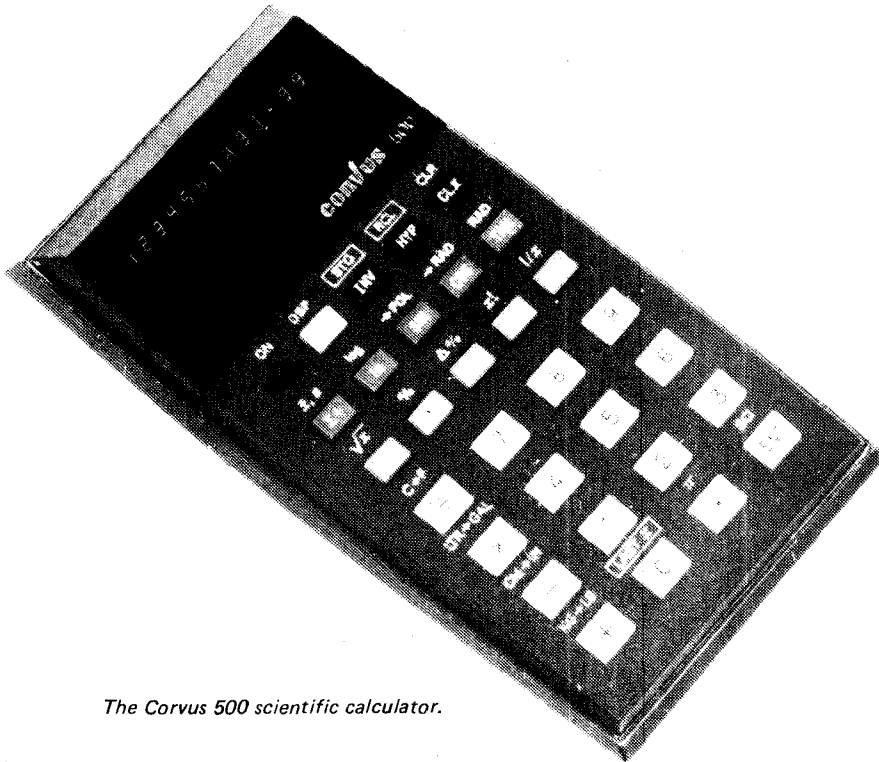
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CALCULATOR OR COMPUTER?

When is a calculator chip a calculator chip and when is a microprocessor chip a microprocessor chip? Rudi Hoess of Electronic Concepts Pty Ltd looks at some calculator ICs that work like microprocessors . . .



The Corvus 500 scientific calculator.

WITH CALCULATORS BECOMING more sophisticated and powerful with every new model, how do LSI manufacturers handle the resultant obsolescence factor and associated costs? Clearly the necessity to remain competitive and constantly upgrade machine capabilities and preprogrammed functions calls for a flexibility akin to microprocessors — where the addition of memory (RAMs) and programs (ROMs) will make upgrades and new model releases easier, quicker and more cost-effective.

An interesting approach was developed by Mostek Corp of Dallas, Texas. The IC combinations described here are used with the Mostek developed Corvus calculators by Electronic Concepts Pty Ltd of Sydney.

Mostek developed an expandable modular family of LSI ICs requiring minimum external components for any given application. A typical chip set centres on the Arithmetic Logic Unit

SCIENTIFIC CALCULATOR = 50103 + 50104 + 50075. This is the combination used in the Corvus 500.

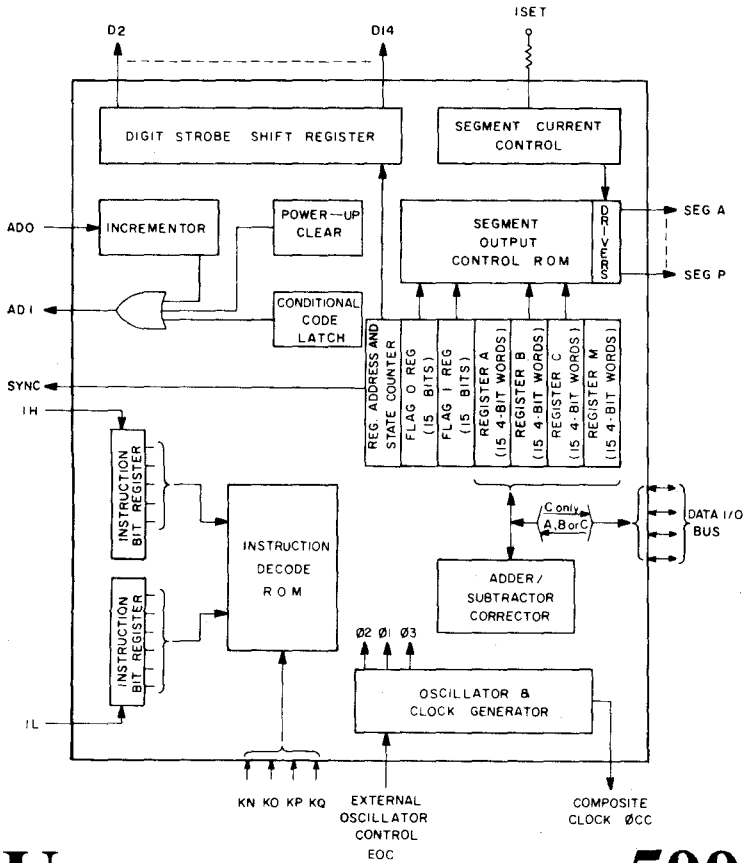
The MK 50075, ALU circuit, combined with the MK 50103, 50104 ROM circuits forms a powerful twelve digit scientific calculator. The display format can be fixed point (user programmed) or floating point in either business or scientific notation. The calculator has four rotatable stack registers plus nine addressable memory registers. All entries use the reverse polish notation.

Effective combination of key functions on this calculator make it possible to offer fifty-five functions with thirty keys. Multifunction keys are accomplished by utilizing the SHFT/DSP, INV/STO, and HYP/RCL control keys. SHFT (Function) enables the upper case key function while INV (Function) enables the inverse of a function. The HYP key is used in conjunction with hyperbolic functions. Several keys have both upper case, and inverse functions while COS, SIN, and, TAN have hyperbolic, upper case, and inverse functions. The order in which the control keys are entered will have no effect on the function. For ex-

ample, both the key sequence SHFT, INV, C \rightarrow F and INV, SHFT, C \rightarrow F will convert degrees Fahrenheit into degrees Celsius.

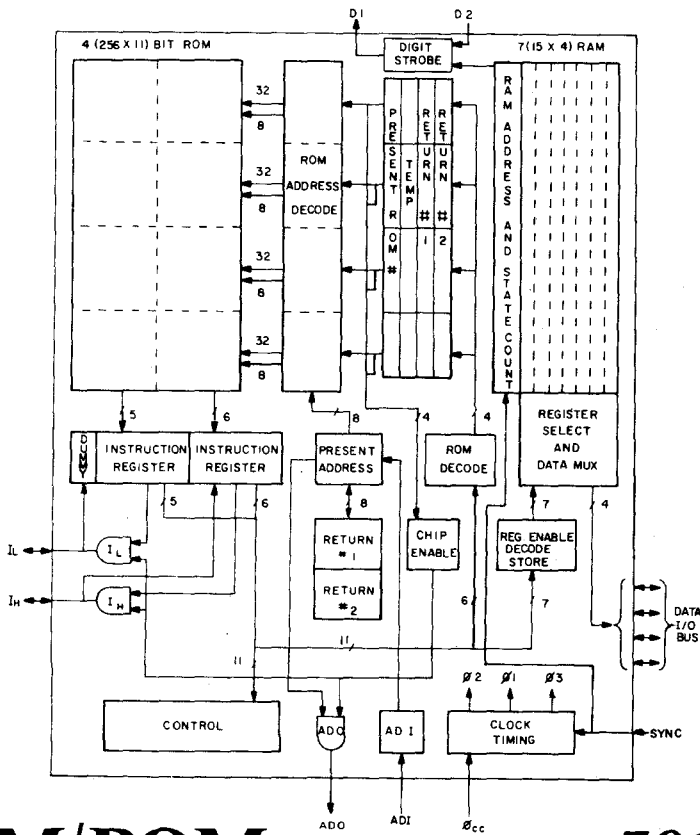
The calculator can work trigonometric functions in either degrees, grads, or radians. When in the radian mode an indicator is turned on. Switching between grads and degrees is achieved by a slide switch while switching from grads to radians or degrees to radians is a key function. The calculator can do the following transcendental functions: $\sin x$, $\arcsin x$, $\cos x$, $\arccos x$, $\tan x$, $\arctan x$, $\sinh x$, $\operatorname{arcsinh} x$, $\cosh x$, $\operatorname{arccosh} x$, $\tanh x$, $\operatorname{arctanh} x$, e^x , $\ln x$, 10^x , and $\log x$.

Besides transcendental functions, it calculates the single variable functions of $X!$, $1/X$, \sqrt{X} , and x^2 and the two variable functions of y^x , $\sqrt[y]{x}$, $\%$, and $\Delta\%$. It has ten preprogrammed conversions which automatically change the contents of the display register into the desired units. It calculates mean and standard deviation using the unbiased method. Its $\Sigma +$, $\Sigma -$, and RCL $\Sigma +$ functions combined with its polar to rectangular conversion makes vector addition straight forward.



ALU

50075

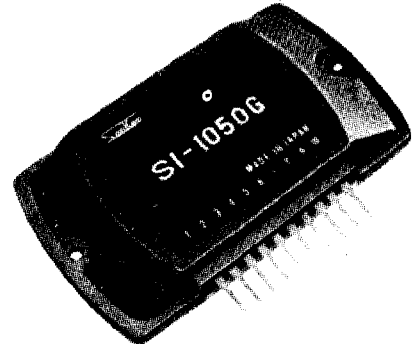


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S1-1030G, S1-1050G

ELECTRICAL CHARACTERISTICS

Characteristic	S1-1010G	S1-1020G
Maximum rms Power	10W	20W
Output Load	8 ohms	8 ohms
Supply Voltage	34V or 17V	46V or 23V
Absolute Max. Supply Voltage	45V or 22.5V	55V or 25V
Supply Current (ave.)	0.50A	0.72A
Protective Fusing	1A Quick Blow	1A Quick Blow
Harmonic Distortion at Full Output	0.5% max.	0.5% max.
Maximum Input Voltage (p-p)	10V	10V
Voltage Gain Full Feedback (P ₀ = 1W)	30dB typ.	30dB typ.

Characteristic	S1-1030G	S1-1050G
Maximum rms Power	30W	50W
Output Load	8 ohms	8 ohms
Supply Voltage	54V or 27V	66V or 33V
Absolute Max. Supply Voltage	60V or 30V	80V or 40V
Supply Current (ave.)	0.86A	1.1A
Protective Fusing	1.5A Quick Blow	2A Quick Blow
Harmonic Distortion at Full Output	0.5% max.	0.5% max.
Maximum Input Voltage (p-p)	10V	15V
Voltage Gain Full Feedback (P ₀ = 1W)	30dB typ.	30dB typ.

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The Fluke 8030A is a new breed of digital multimeter.

It's not like earlier DMM's that were small, but had nothing to offer. Instead, the 8030A has all the performance, features, accuracy and resolution you would expect from a benchtop instrument at a much higher price.

It's an inexpensive, handheld 3½-digit DMM. But it's also built to Fluke standards and Fluke quality. That makes the difference.

The 8030A DMM is designed for use by troubleshooters and field service technicians. It offers six measurement functions in 26 ranges. Automatic polarity and automatic zero suppression are standard. Rugged construction. Small size. Accuracy specifications that are guaranteed for a year. Plenty of accessories. Even a built-in hood that pulls forward for easier reading in sunlight.

But the most important news about the 8030A is all the features we've designed into it.

Measurement

The six measurement functions are ac volts (true rms), dc volts, ac current (true rms), dc current, resistance, and diode test. Each function has five selectable ranges, except diode test which has one.

True RMS

True rms conversion techniques provide ac measurement accuracy (voltage and current) in spite of noise levels and

distorted wave forms. AC coupling rejects dc bias during ac voltage measurements. And dc coupling on ac current ranges provides ac + dc capability for measuring power supply and SCR regulating circuitry.

In-Circuit Resistance

In circuit resistance measurements can be made using all five resistance ranges. Errors due to effects of parallel semiconductor junctions are avoided by applying a maximum voltage of 200 mV across the on-scale unknown resistance.

Diode Test

The diode test function allows measurement of diode and transistor forward-voltage drops. A 1 mA bias current is forced through the junction, and the resultant voltage drop is displayed in millivolts. This function can be used for in-circuit testing of transistors and diodes and will detect up to three junctions in series.

Overload Protection

Extensive overload protection is provided, and, in addition, units have successfully withstood transients up to 6,000V across the input.

Functional Self-Check

Each of the six measurement functions can be checked for basic functional operation by using regular instrument test connections. No external test equipment is required.

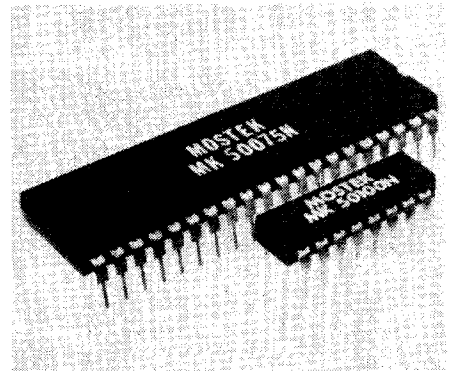
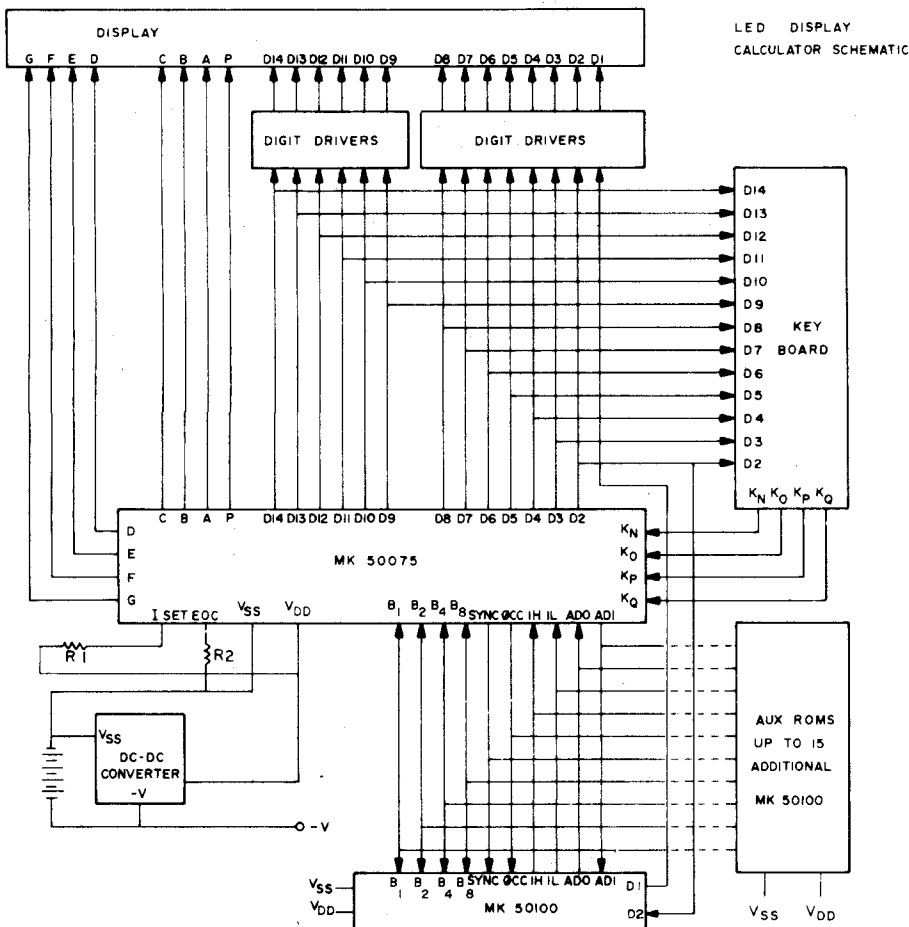


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CALCULATOR OR COMPUTER?



MK 50075, designed to handle all typical calculator in and output requirements, directly interfaced to a RAM/ROM/Interface chip MK 50100 containing 1K words of instructions (ROM), seven 15 digit registers of read/write storage (RAM), plus all I/O interface requirements to facilitate automatic synchronisation with the ALU.

Maximum configuration allows for 16 MK 50100 memories to be connected to one 50075 ALU.

To develop a new application, or expand an existing one, only the ROM area needs to be imposed on a standard MK 50100 — thus simplifying the creation of a new product or the upgrading of an existing product without the development costs inherent in the creation of a ROM of similar capability and ease of interface.

SYSTEM

BUSINESS CALCULATOR = 50101 + 50102 + 50075. This is the combination used in the Corvus 600.

The MK 50075, ALU circuit combined with the MK 50101, MK 50102 ROM circuits forms a nine or twelve digit algebraic business calculator. The calculator normally has a nine digit display although all internal calculations are computed to thirteen digits. By adding two diodes and a switch twelve digits can be displayed. The display format may be fixed point (user programmed) or floating point.

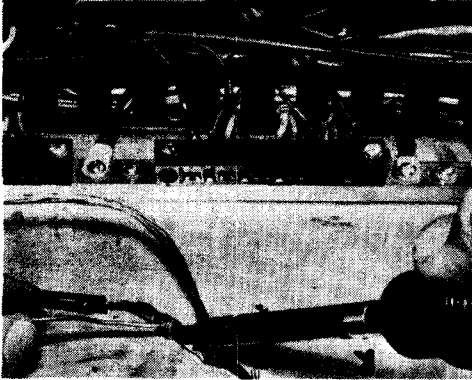
The chip set has nine memory registers. The memory operations of addition and subtraction are performed on the contents of memory register one and the display register without affecting the display contents.

The calculator has the standard four functions of addition, subtraction, multiplication, and division. The equal key serves two functions. The first time it is

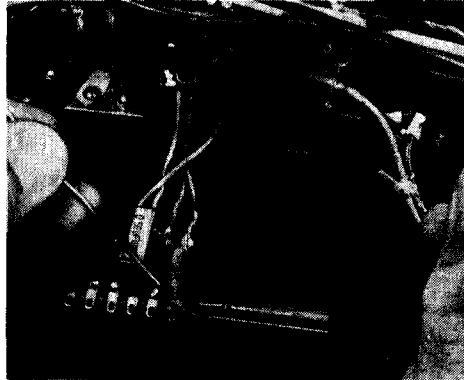
pressed, it completes the previously entered operation. The second time it is pressed the calculator will perform constant operations in which the second numerical entry is treated as the constant. The calculator also performs the functions of percent, y^x , and gross profit margin.

Stored in its ROMs, this chip set has a unique financial program. Tied in with this program are five financial keys: i, interest; n, number of payments; PMT, amount of payment; PV, present value; and FV, future value. All financial calculations require a minimum of three input parameters to solve for the other two. Input parameters may be entered in any order from the keyboard or be stored in the financial registers as the result of previous calculations. These keys make financial calculations easier than adding or subtracting.

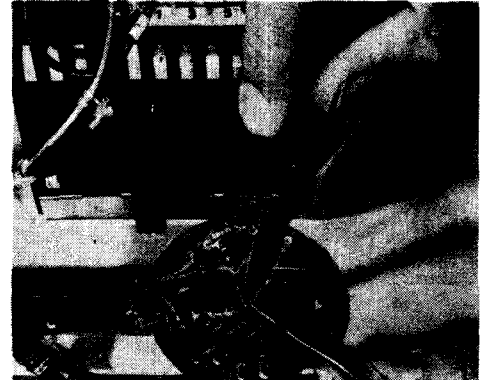
"WITHOUT MY SCOPE IRONS I RECKON I'D NEED A 25 HR. DAY."



8.00am "Start the day with a heavy earth connection on the emergency power plant. Need a 130 watt iron. My Superspeed's got that and more. Just as well, the workshop's 400 yards away".



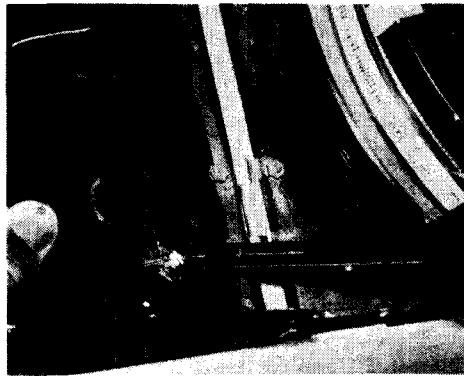
9.15am "Call from No. 2 Moulding Shop. Ran some temporary leads involving both tag joints and Printed Circuit Board (P.C.B.) connections. The Scope Minispeed handled the lot".



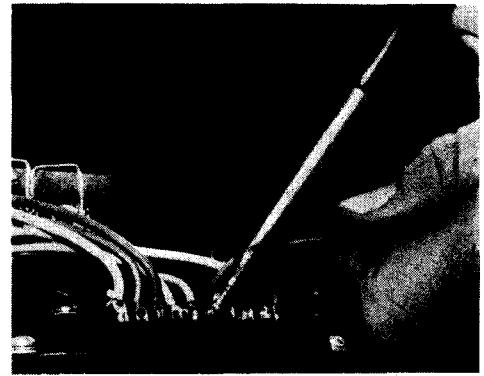
10.55am The V.D.U. in the process control room is playing up. Replacing miniature resistors on a P.C.B. is, ideally, the job for a temperature controlled 60w iron but my Scope Minispeed did the job and did it well.



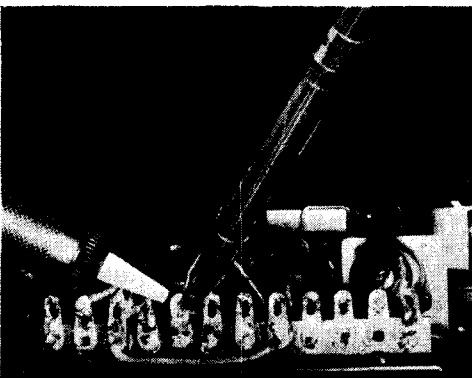
11.15am "Fix the speaker leads in the canteen P.A. Need a 30-40 watt iron, but my Scope Minispeed did the job".



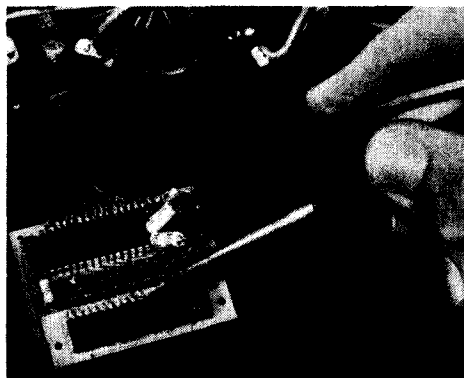
1.30pm "Resolder a 7/036 earth to sheet metal — LP gas flame would work, but too much risk of heat damage to PVC cable. The Superspeed iron produced its full 150 watts and did the job".



2.15pm "Fred borrowed my Minispeed to tackle an open circuit on the fork lift's headlights. He permanently soldered the wires to the terminal block, and used the Mini's 75 watts to do it".



2.17pm "Tag soldering in the workshop and a desolder job on a P.C.B. Using the Minispeed saves swapping between conventional 60 watt and 25 watt irons".



3.58pm "Emergency in shop six. I used my Minispeed to unsweat the leads of the main heater control circuit and then desolder the pyrometer circuit on the P.C.B. One iron, two different jobs".



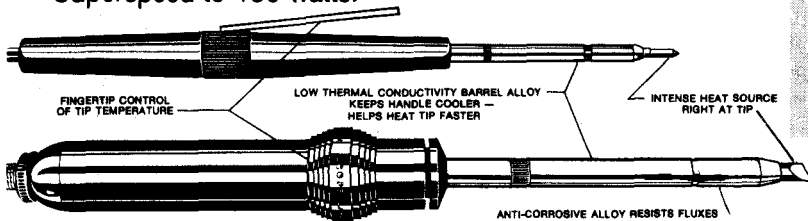
4.18pm "This wiring's a real birds nest. Passed through the wires with the Minispeed stone cold, desoldered the three joints, let the iron cool down, then withdrew through the PVC insulation. The 5 second heat up and low tip mass let's me do this".

Scope soldering irons save time three important ways:

1. Versatility.

One Scope iron replaces several conventional irons because it can tackle a wide range of soldering problems, from integrated circuits and printed circuit boards to heavy earth and chassis connections.

You don't have to swap irons half-way through the job. Both the Minispeed and the Superspeed function as 20-30 watt irons, and then within seconds and a touch of the finger switch, you get increased heat output to increase the Minispeed to 75 watts, and the Superspeed to 150 watts.



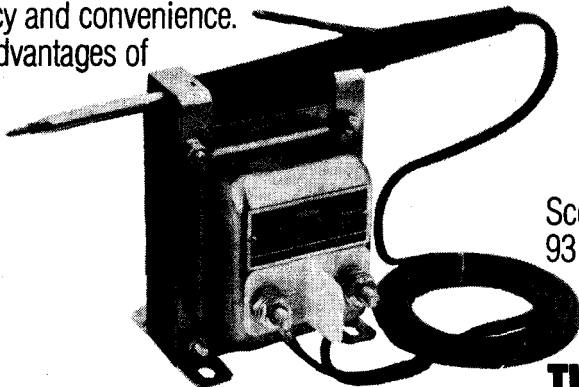
2. Speed.

Five seconds gets both irons ready for work, and they cool down quickly as well. When you encounter a heat sink you want an iron that can make up the heat loss instantly and maintain good soldering temperature. Result, the job's done fast and you can move onto the next.

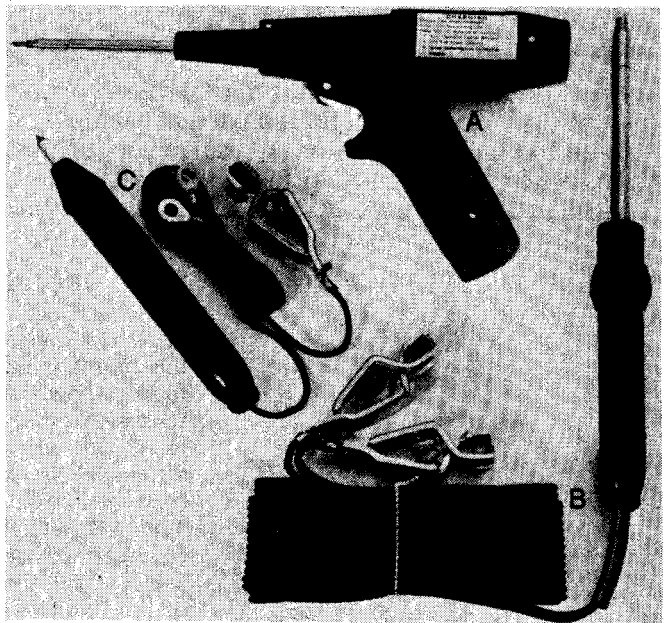
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Scope irons are isolated from the mains. The special isolation transformer lets you work on live equipment with a higher degree of safety than a conventional iron.

The Scope range of products is designed to deliver efficiency and convenience. Consider the advantages of these products.



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(B) Scope 12V Hobby Iron.

This versatile iron is designed to work within 6 metres of your car battery.

(C) Scope Vibroscope.

This electric pencil allows for permanent writing on all metals. Valuable in an engineering store identifying metal tools, dating and naming parts, inscribing trophies.

For enquiries and further information on the Scope range of products contact: Scope Laboratories, 93 Matthews Avenue, Airport West, Melbourne, 3042.

SCOPE
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COMPUTER APPLICATION REPORT

Well we've talked enough about microprocessor evaluation kits and how to get them going . . . but what about an actual example of a company buying one of these and putting it to work — using a microcomputer to test the car burglar alarms which are on special offer in this issue.

In this photo you can see six alarms plugged into sockets on a piece of board. In all there are sixteen alarms tested at the same time.

The commands are entered via the keyboard, but once the program was written this was used only to start the test. The printer on the right gives the results of the test.

The middle box in the stack of three in the centre of the picture holds the Motorola evaluation board and the interface board. These can be seen in the photographs on page 32.

THE PROBLEM: TESTING A 'Sleeping Dog' burglar alarm normally takes one person about ten minutes, this means it would take three months to test a few thousand, but it is needed to test at the rate of this many per week. One way to do this is to employ sixteen or seventeen workers, but a better way is to use a microcomputer.

THE SOLUTION: By using a microcomputer the desired testing rates were achieved and there was the added bonus of an individual printout giving the specifications of each alarm. And human error was eliminated.

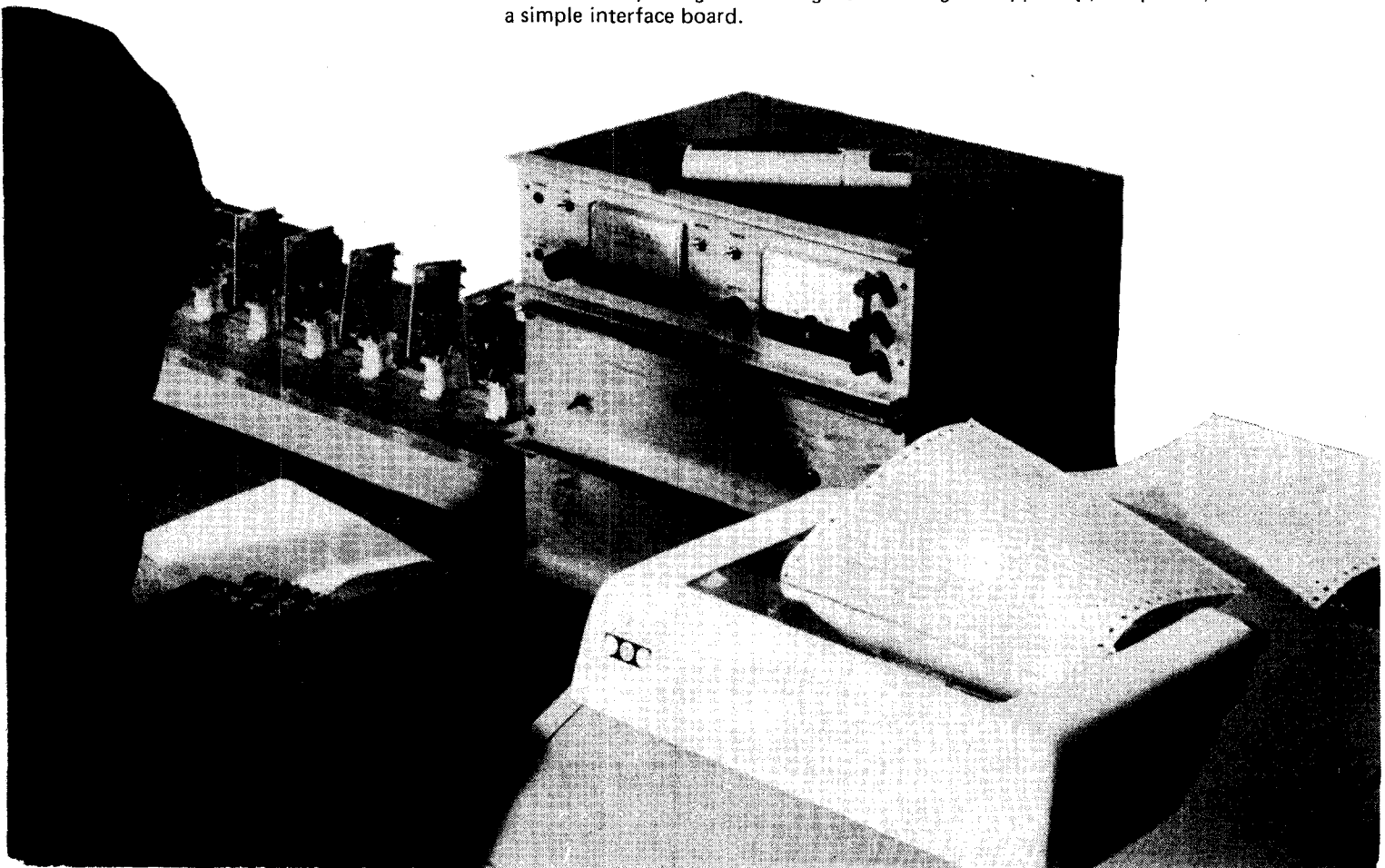
The system developed by the company enables alarms to be tested 16 at a time and apart from program development the only thing to be designed was a simple interface board.

Testing Procedure:

1 To test reverse-voltage and over-voltage protection circuits, the alarm is deliberately abused in these two ways before the testing starts.

2 To test arming of the alarm, the voltage is removed from the 'ignition' input (to simulate switching off the ignition) and the signal on the 'lamp' output is monitored. An intermittent voltage (to flash the lamp) should appear after 90 seconds (this indicates that the alarm is armed).

3 To test the trigger input (connected to the car door) and the horn output, the alarm is triggered and after 10 seconds an intermittent voltage should appear on the horn output (to make it go beep, bleep, beep . . .). After two



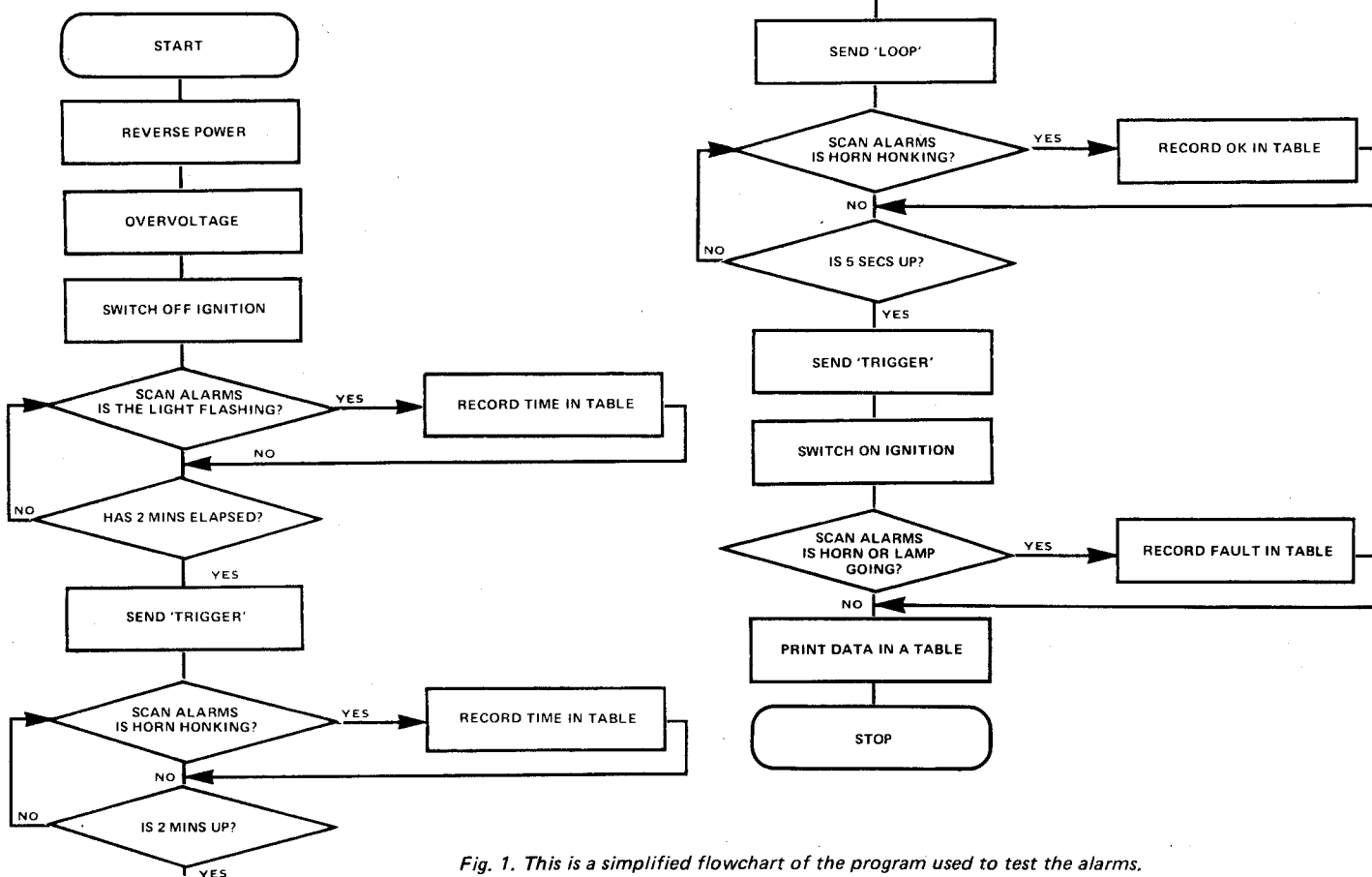


Fig. 1. This is a simplified flowchart of the program used to test the alarms.

minutes the horn should stop and the alarm remain armed.

4 To test the 'loop' input (which protects accessories on the car) the normally-closed contacts across the input are opened and the 'horn' output is monitored to check for two minutes of beeping.

5 To check that the alarm switches from its armed to its dormant state when the ignition is switched on, the alarm is triggered and then the 'ignition' input is taken high within ten seconds. The absence of outputs indicates that the alarm is dormant.

The testing procedure is the same

```

TEST RESULTS, ALARM # 0101
EXIT TIME 74 SECS
ENTRY TIME 18 SECS
TRIP ALARM 66 SECS
LOOP ALARM 64 SECS

```

Fig. 2. The test results for alarm number 101.

for the computer as it is for the human tester. The flow-chart shows, in simplified form, how these tests are carried out on sixteen alarms at the same time.

How The Computer Does It

Looking at the flow-chart you will see that the program can be broken down into a series of logic functions. As it starts its program it needs only to put a suitable bit in a latch on the interface board and relay logic takes over to reverse the power supply to all the alarms. In a similar way the computer can provide the overvoltage instruction, the ignition-on and ignition-off voltages, the closed and open conditions on the loop circuit, and the trigger voltage which simulates opening the car door. These five computer outputs (alarm inputs) are decoded from a 6-bit bus on the interface board. The status of the lines in this bus results from the data put into certain memory locations on the processor. On the interface board a decoder looks at these bits and puts five relays into appropriate states.

This takes care of the signals from the computer to the alarms (all 16 alarms are spoken to at the same time) and before we look at communication in the other direction we will look at the need for isolation and how the interface board is organized.

Isolation

Above we mentioned that relays are used to isolate the computer circuit from the alarm circuit. Isolation is needed because the alarms work on a 12 V supply and the computer uses 5 V, also noise immunity is improved. When isolating the two circuits the designer had the choice of opto-isolators or relays. He chose relays because these can easily be protected against high currents (by a simple fuse). An opto-coupler would probably blow up when faced with short-circuited alarm.

The Interface Board

We have looked at the way the computer sends out signals to the alarms via relays. Now we will see how it reads the outputs from the alarms. Each of the sixteen alarms has two outputs (one to flash the lamp, one to beep the horn). There are simple networks on the interface board to differentiate then integrate these intermittent signals, so a steady high or low voltage will not get through.

These thirty-two outputs are multiplexed to a single alarm output line. When the flow-chart says 'scan the alarms to see if there is a flashing lamp signal' the computer looks at this single output line and addresses each of the alarms in turn. As there are 32

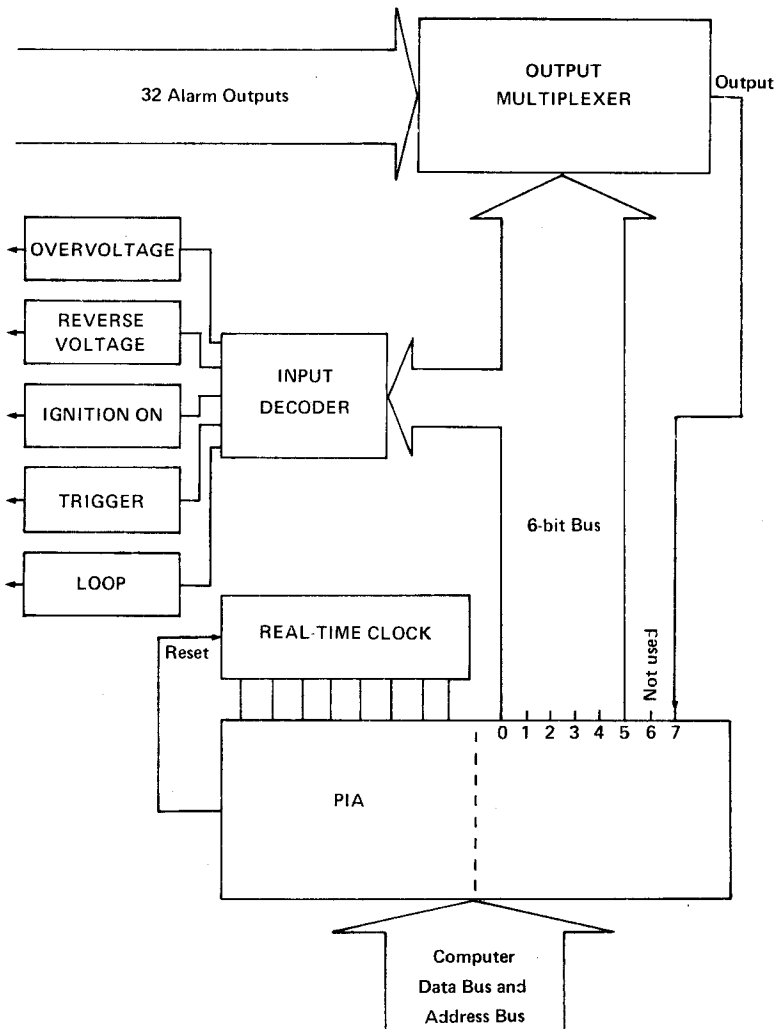
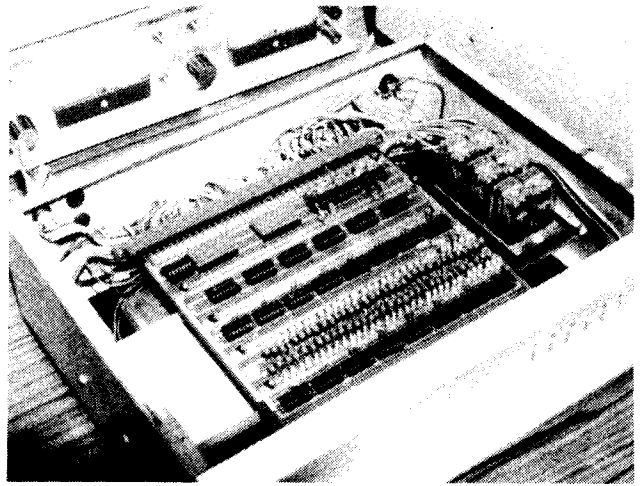
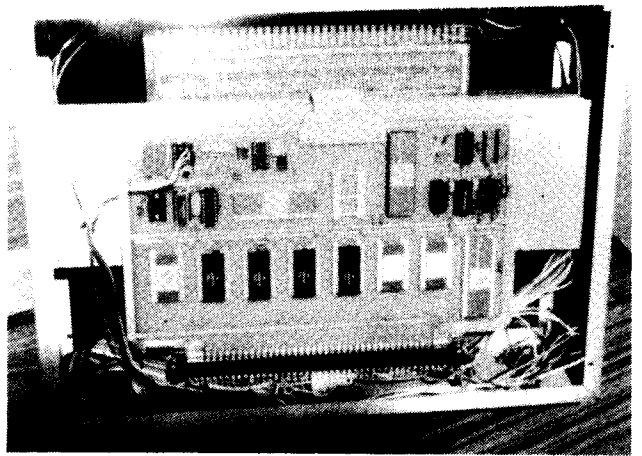


Fig. 3. Block diagram of the interface board.



This photo shows the interface board mounted in the top of the computer housing.



This photo shows the Motorola evaluation kit mounted in the bottom of the computer housing.

outputs each can be uniquely addressed with a 5-bit code from the computer. This code is sent along the 6-bit bus from the PIA (see the diagram). The outputs have addresses from 00000 to 01111 (note the first bit is always 0). When the computer tells the PIA to send out a code starting with a 1 this is decoded by the input decoder and the appropriate input signal is sent to the alarms.

The interface board contains the 6-bit bus, the logic to decode addressed outputs and inputs, the relays and interface components to handle signals to and from these inputs and outputs, plus two other sections. These are the PIA and the real-time clock. We will look at the PIA in more detail later; for now it can be regarded as two 8-bit ports to the computer. One of these is an input port, the other is used as a 6-bit output port (driving the 6-bit bus) and a 1-bit input port (one pin is unused).

The Real-Time Clock

The 8-bit input port in the PIA is connected to a real-time clock. This counts seconds in binary and is available for the computer to read at any time. The reason an external clock is used (rather than telling the computer to count cycles) is because of the magnitude of the times involved (a computer cycle is around 2 microseconds). The clock is reset by the computer with a signal on a peripheral control line from the PIA.

The PIA

This IC sits on the data and address buses of the computer and has two sets of 8 lines which can be used as inputs or outputs in any combination. This setting-up is done by writing prescribed numbers into certain memory addresses. Each of the 8-bit ports has a memory

address through which the processor can insert or extract data.

The System

The system used in this application was based on a Motorola evaluation kit. In addition to this board and the interface board described above all that was used was a printer (this is not essential for simply testing the alarms; its inclusion in the system is simply to provide diagnostic data to assist fault-finding), a power supply and the hardware for connecting up the alarms. The program was contained within the RAM of the system in the photographs but in the long term it will be included as a PROM.

For developing the program an alphanumeric keyboard was used (this came ready-encoded to give ASCII output and a UART was added so that the monitor program in the evaluation kit would see it as a teletype). ●

WHERE THERE'S A NEED THERE'S A STANTON



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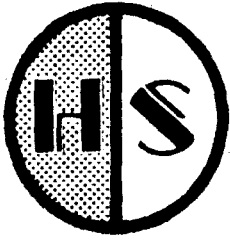
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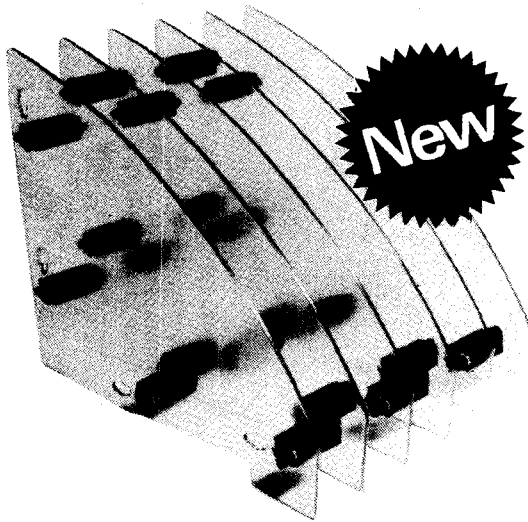
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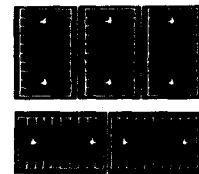
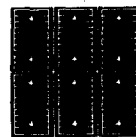
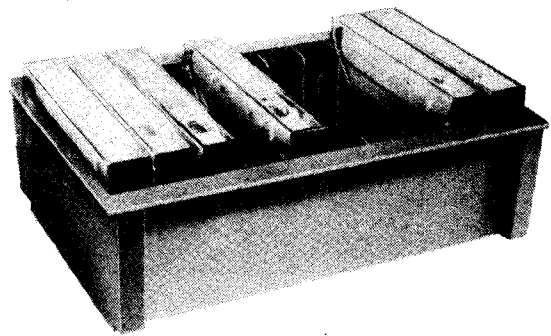
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The alarm can also be activated by breaking a loop to ground connection, in this case, activation of the horn is instantaneous. The ground loop may have a number of switches connected in tandem and finally to an earth point such as a radio, trailer or cassette deck. The earth loop may also pass into outside mounted lamps — when the wires to the lamp are cut, the loop is broken and the alarm activated.

When activated, the car horn will sound intermittently for two minutes after which time it will stop and the alarm will be reset.

A lamp which may be mounted on the dash flashes while the alarm is set, as a deterrent to would-be burglars.

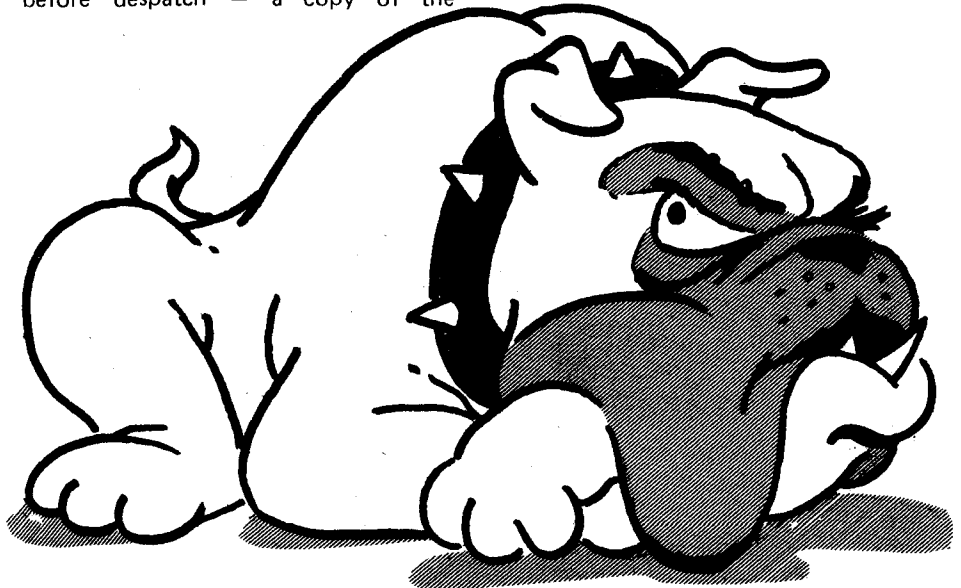
When the ignition of the car is switched off, there is a 90 second delay before the alarm is armed allowing reasonable exit time.

the time depending upon the model of car and how well one knows its electrical system.

The 'Dog' is extremely well designed and made — an example of Fairlight's thoroughness is that they have built a microprocessor-based testing system and not only is every single alarm tested before despatch — a copy of the

relevant computer print out is supplied with the unit!

Electronics Today International has no hesitation in thoroughly recommending this unit. At its normal selling price of \$58.50 it's very good value — at the special offer price of \$39.40 it's even better.



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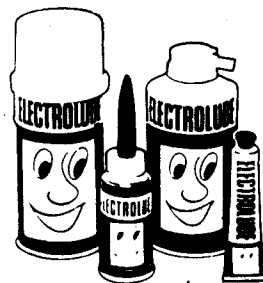
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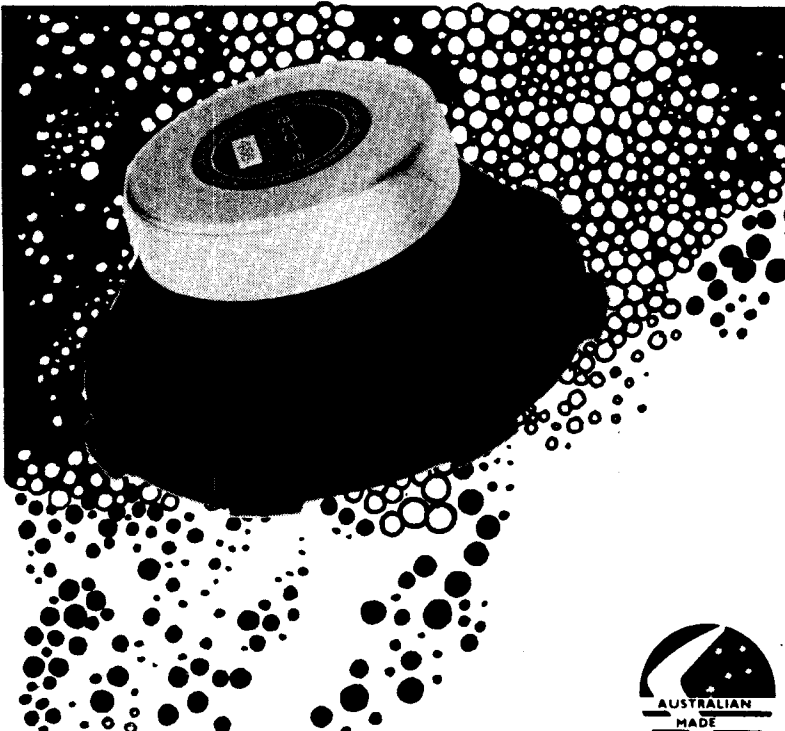


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AD83

ELECTRONICS

IN THE YEAR
2000

EVOLUTION OR REVOLUTION

By Peter Sydenham

SCIENCE PROVIDES US WITH knowledge about existence. It is based on a procedure of collecting facts which are placed into apparently logical groupings in order to lead to stage two of scientific method – the realisation of one or more hypotheses. Man's imaginative powers then enable ideas about these facts to be "tried out" in the mind. (The mind creates what are called 'models'). After a brain-storming session some ideas emerge about the collected facts. These are likely contenders of generalised models that will describe many seemingly different ideas by one unified concept. Figure 1 depicts this process.

Having hit upon an hypothesis it is then tested by performing experiments upon it to see if more examples that would appear to also be correct are

indeed allowed for. The hypothesis, as long as it is found satisfactory, is then held as current and applicable until a new case emerges that is not described adequately enough by it. The scientific process is then begun again to find a new hypothesis that is better than the earlier one.

Old hypotheses are not necessarily useless. They find their use in limited cases. We are quite satisfied in everyday life to regard mass as a constant entity but on some special occasions, in the design of some cathode ray devices for example, mass must be considered as being convertible to energy. Einstein's work predicted that conversion process.

New hypotheses produce new ideas for technology to take up and apply. Once it was *known* that the atom was

divisible, scientists sought to split it further.

This brings us to the role of technology in the development of ideas. Technology and engineering is the broad discipline that devises machines and structures that do not exist as such in nature but using resources that are available naturally – see Fig. 2. Machines provide us with power conversion, with mechanisms and with measuring and information tools.

Technology is a sister requirement of scientific pursuit – inseparable partners in progress, each affecting the other's progress at varying degrees with time. Figure 3 shows an example of this interaction.

One often-seen mis-statement is that scientists build the so-called scientific machines. "Scientists put a package on

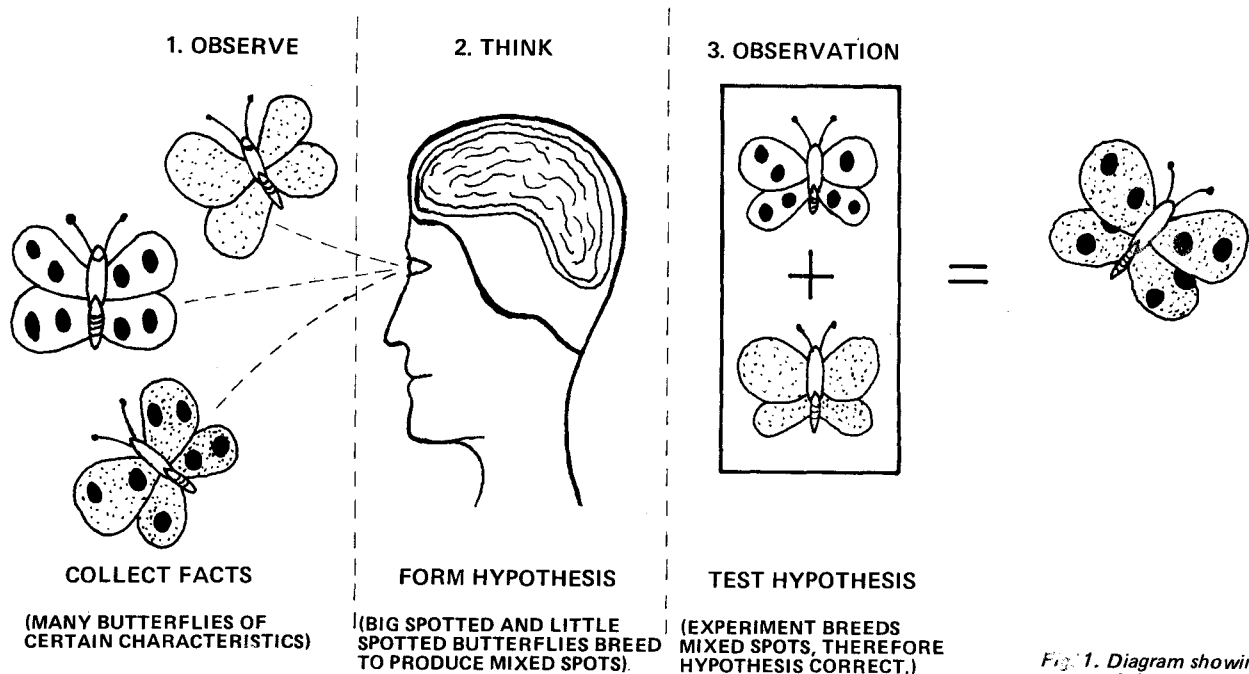


Fig. 1. Diagram showing stages of the scientific method. Science produces knowledge.

Electronics 2000

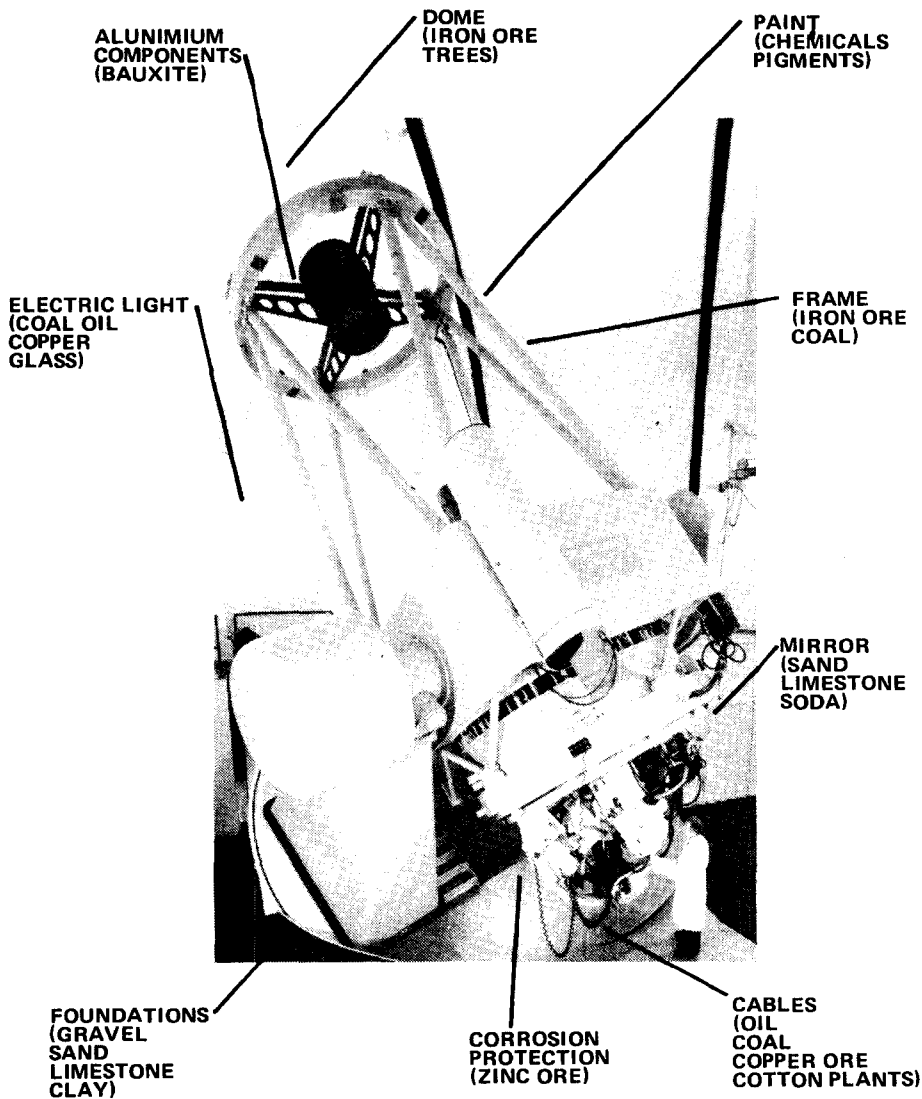


Fig. 2. Technology produces new objects, using naturally available resources, for specific tasks. (98 inch telescope at Royal Greenwich Observatory).

pared to accept change and make what appears at the time to be sacrifice more readily. We have seen over the past two years a strong swing toward the smaller car. Air travel is now cheaper than by sea — the reverse of a decade ago. A trans-world telephone call now gives more message content than a telegram for the same price.

This introduction and the previous part of this study sets the scene for what I see could be some aspects of future living. I possess no crystal ball; I claim no extra sensory perception ability, nor do I have a pact with the maker or devil! What is given now is composed from studying the past trends and extending them into the future, this being sprinkled with some personal ideas of myself and others.

GETTING ABOUT IN THE FUTURE

Although there have been instances in history where knowledge of man has been lost by chance or by political decree (the 1930's burning of the books in Germany) — technological change has continued to advance in all civilisations (albeit sometimes extremely slowly). It is most unlikely that the "alternative" communes we see today will be how man will live in future. It would need a global catastrophe to destroy all technology so thoroughly that the survivors would have to live as cave men and reinvent all inventions again.

Technology of long-distance transportation — railways, ships and even cars — requires great financial investment. Few people can afford a hand-made car today — even mass-produced ones are becoming harder to reach. Thus, if big commercial business survives into the 2000s, or the State takes over, we can confidently expect changes to slowly emerge in transport.

It takes about four to six years for a current design railway and its rolling stock to be built from conception. A new technology such as airships (a revival really), needs a decade and a half once a serious commitment is given to using it.

Electric cars are constantly being researched and developed — Fig. 4 — but it has become vitally clear that two areas of difficulty exist. The first is that the lead-acid battery is not adequate to power a car with performance that we have become accustomed to. The best produced to date is not an equal to the smallest family petrol car. What is needed is much more effective

Mars is the greatest scientific achievement yet made by man". If it had been a failure then it would have been due to engineering failure! The Mars' probe is rather the greatest *technological* achievement.

It is important to see how much technology compared with how much science goes into a manufactured product for this helps us predict when new ideas will come into practical use.

There is, however, another aspect to technology. Many lifestyle changing ideas do not occur as the result of applying science in a systematic manner. In fact many valuable machines and ideas arrive by way of an unknown,

often poorly trained, inventor who applies an uncommonly good amount of common-sense to solving an immediate problem.

But society itself is also a strong influence on the application of new ideas. Somehow a new idea appears out of place. We now accept spectacles as normal technology, but think how a person wearing a filtering false nose (a possibility for reducing hay-fever allergies) would be received. Organ transplants were, and still are to some extent, opposed. Test-tube babies are currently controversial.

Many problems of society could be solved more speedily if we were pre-

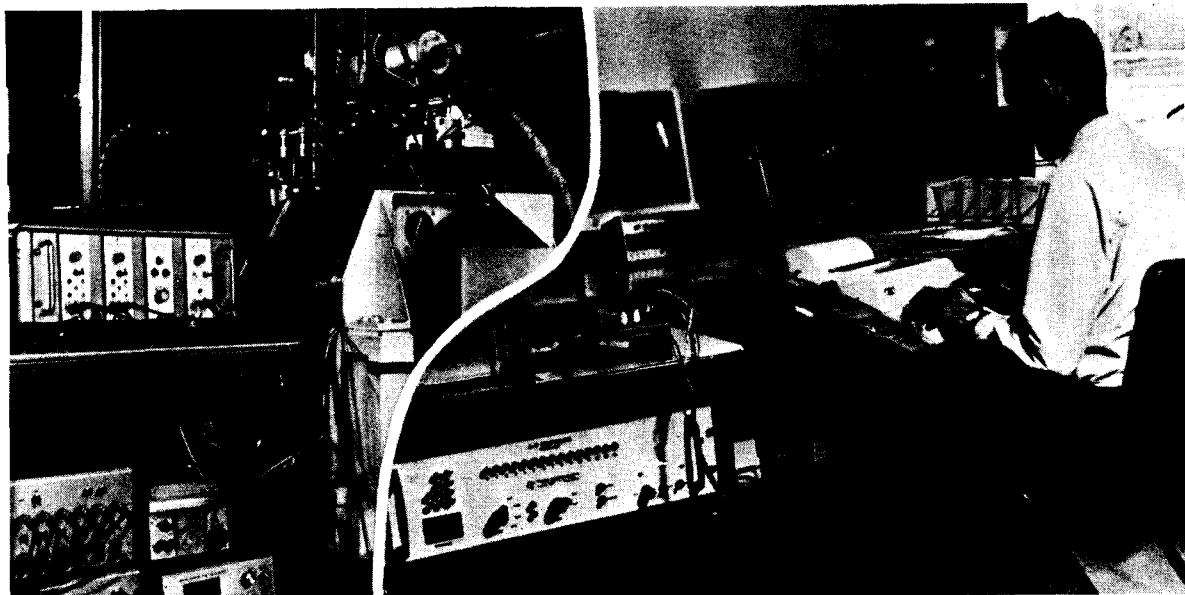


Fig. 3. Science and technology go hand in hand helping each other continuously. (Computer controlled engine testing at Cranfield, UK).

electrical storage arrangement. High-temperature sodium-sulphur batteries, as pictured in Fig. 5, have been developed to prototype stage. (See ETI, December, 1974) but the manufacturers suggest these cells will not be available to car makers before 1980. Add a few years more for design testing in practice and we may have a better electric car proposition by 1985. The second difficulty, however, may be a more serious barrier to the widespread use of electric vehicles. This is the general reluctance to move on a downward trend of performance. Petrol will still be available in the '90s so the choice will probably move toward smaller, just as well-performing, cars that stretch the litre further.

Other likely developments are automatically steered vehicles running on specially modified highways. The cars would be guided by control units sensing guidance paths laid in the surface. Collision prevention by short distance doppler-radar and optimal route selection using telemetry signals picked up by radio or from transmitters also laid in the road (ETI, September, 1974) can be implemented now that research is in progress on these devices.

Possible schemes are shown in Fig. 6. The computing capability needed, including built-in redundancy to improve reliability, is now available in micro-processors that will soon be as cheap as a good transistor radio. Social influences, people's suspicions and mistrust and overall cost are the constraints on rapid developments in this area.

With the thirst for speed perhaps settled to a reasonable level the next thrust will be safety and again, perhaps,

longevity of the vehicle. New ideas obtain much publicity — but promotion and worthwhileness of the product are not always related attributes.

Experimental Safety Vehicles, ESVs, have reached advanced levels. Figure 8 is Nissan's E2 model. Urethane bumpers are now being used; other safety features are gradually being introduced. Perhaps vehicles adhering to the surface will be displaced by slightly levitated ground-effect machines like the hovercraft. Again, experience has shown that these are not the complete replacement for all ground transport systems. As yet they are still noisy, power hungry and not as responsive to directional control as the wheel-borne car.

The electric tram is another develop-

ment that may come back again in a new form. Melbourne, for example, has a workable mass passenger transit system that now gets people into and out of the city generally faster than by car — yet it was not so long ago that the tram was regarded as archaic. Today it is recognised that mass transportation routes are better for concentrated city mobilisation than a melee of cars.

Magnetically levitated vehicles running on relatively inexpensive tracks were forecast to be capable of over 500 km/h speeds (see ETI July, 1973). Development of short test tracks and vehicles continues but the pace of development is slow to provide economic alternatives to maintenance and repair of existing systems. Figure 9



Fig. 4. The electric car can only provide satisfactory transportation for short haul tasks. This is a pleasure version by ElecTraction.

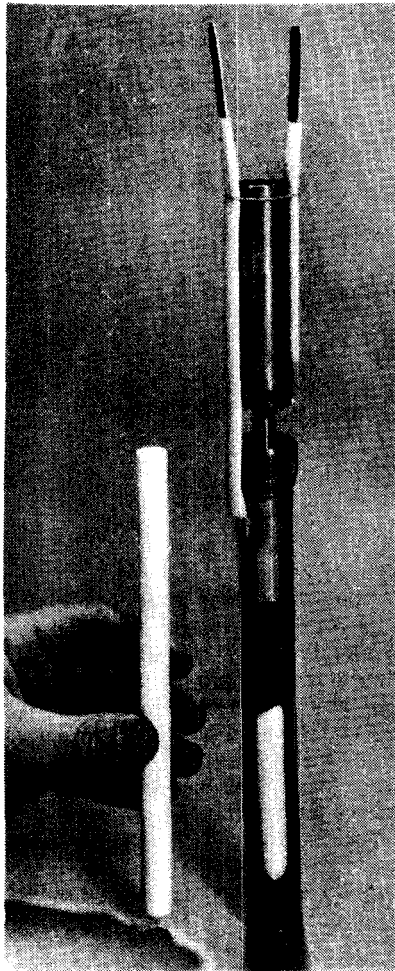


Fig. 5 Na-S high temperature batteries can store much more power for a given weight than lead acid cells but they will not be available for another five years or more.

shows Maglev vehicles around 1973 — remember that prototypes are quicker to materialise than service vehicles. For inter-city distances in big countries we need a speed of about 500 km/h to make journeys sufficiently shorter than the current alternative of the car or train. Airflight time is becoming limited by cost and airport turnaround time — door to door and with no connections to make, a 1000 km distance takes about three hours. Maglev inter-city systems, however, now must be designed in light of a new social barrier not obviously in existence five years back — they need great quantities of power to run at such high speeds. Societal values no longer ignore such demands on resources. Superconducting Maglev systems (ETI, June, 1973) will require vast quantities of scarce helium — this may limit their widespread useage.

The bicycle is good for the health

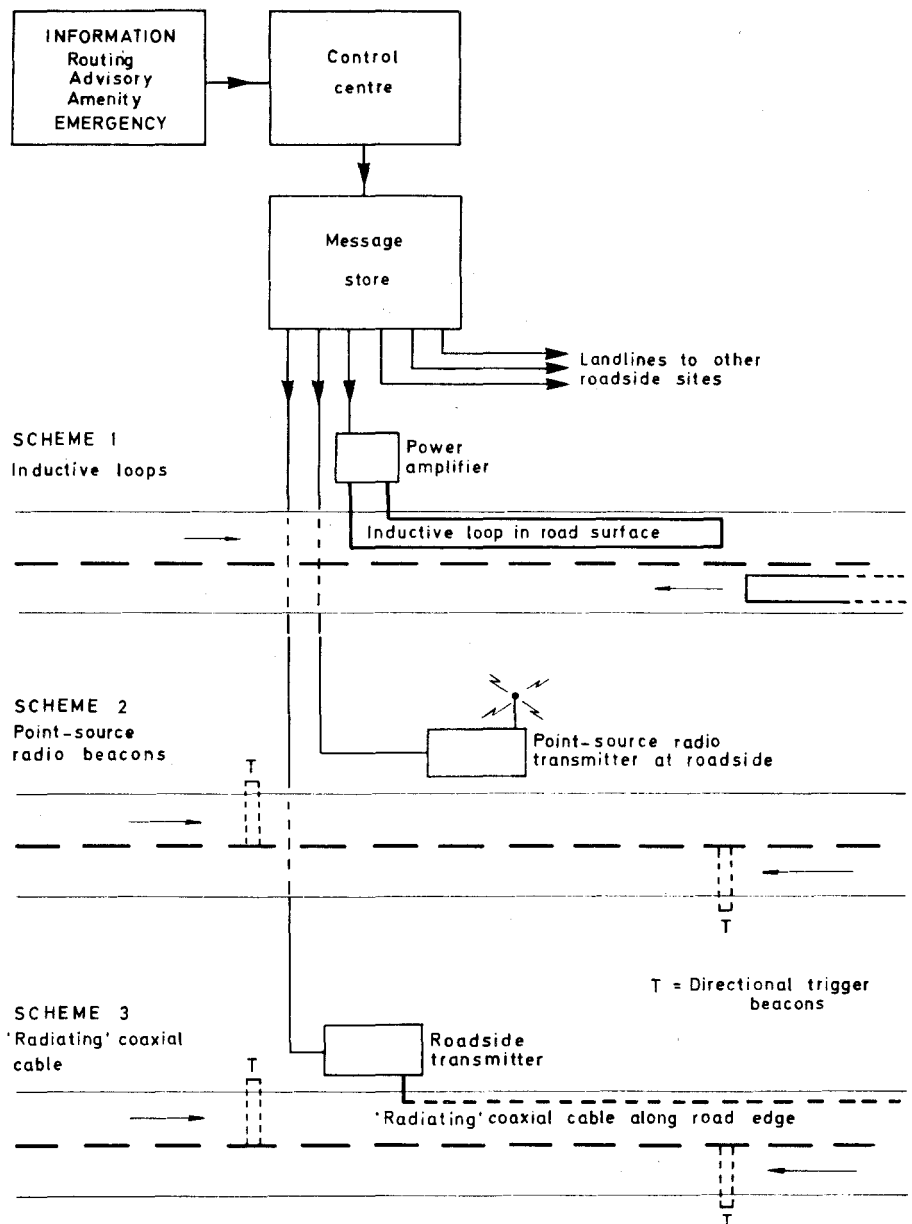


Fig. 6. Alternative methods by which drivers may be fed information about their optimal route and safety conditions ahead.

but its slowness and effort requirements do not suit most people who live far from their workplace. A compromise between the bicycle and the car seems how things should develop but social constants, personal comfort and the ability to carry passengers and loads require a wide degree of flexibility for the future personal transport vehicle.

Moving on to transport at sea we can confidently expect to see automatic ships navigating by electronic control. Position sensing devices (see ETI, December 1973 and April 1974) are sufficiently developed for the task —

especially when the Omega navigational net is complete across the globe. Computer control is well capable of the data processing needed and machinery control is now extremely reliable and well defined. Automation of ships, however, would need global acceptance of the concept and more faith in machinery. Automatic fishing is also a realisable goal — for we can now detect where fish are in the sea.

Ship-forms may shift to surface-effect designs — plans were announced in the US last year for design studies of a 10 000 tonnes naval cruiser that

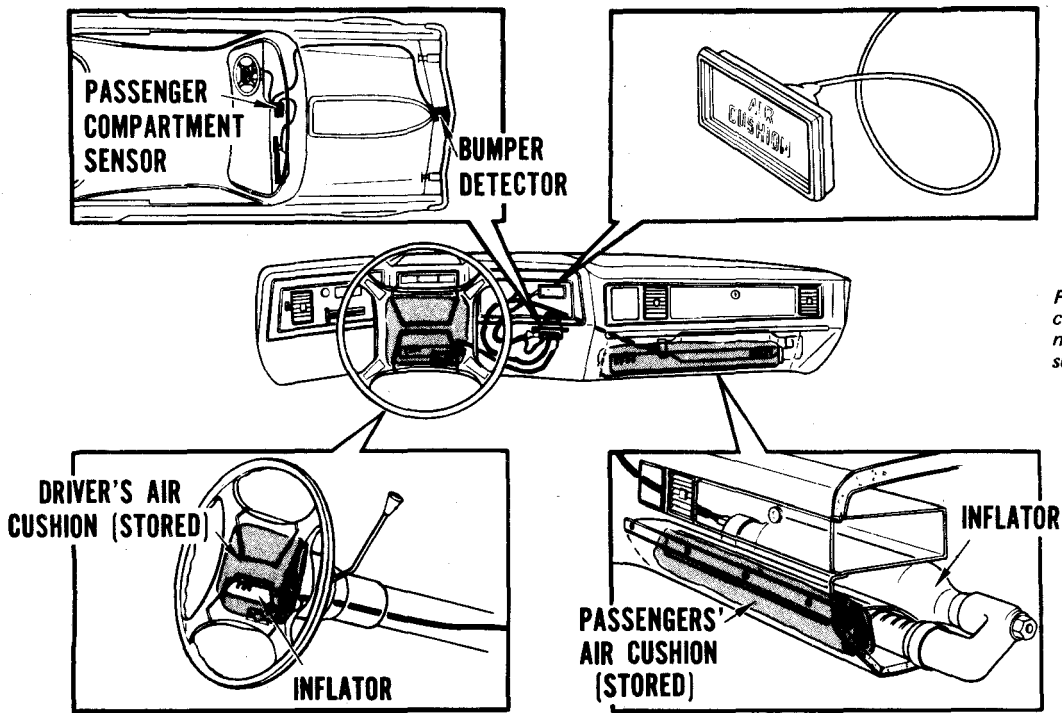


Fig. 7 General Motors' air cushion restraint system. These may be the answer to collision survival in later years.

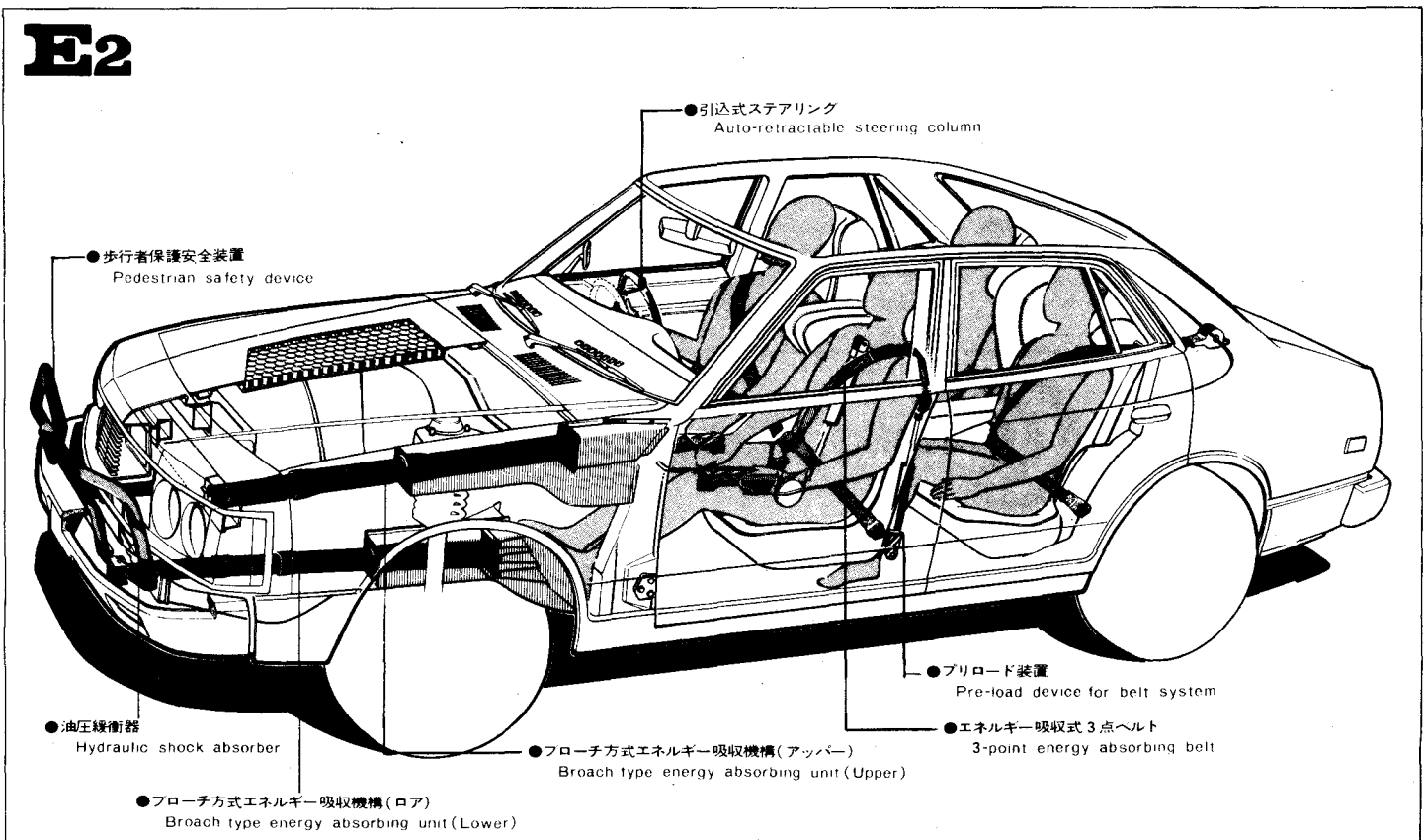


Fig. 8. Nissan E2 type ESV incorporates ideas yet to be included in production cars. Gradually changes to improve safety are being incorporated.

Electronics 2000

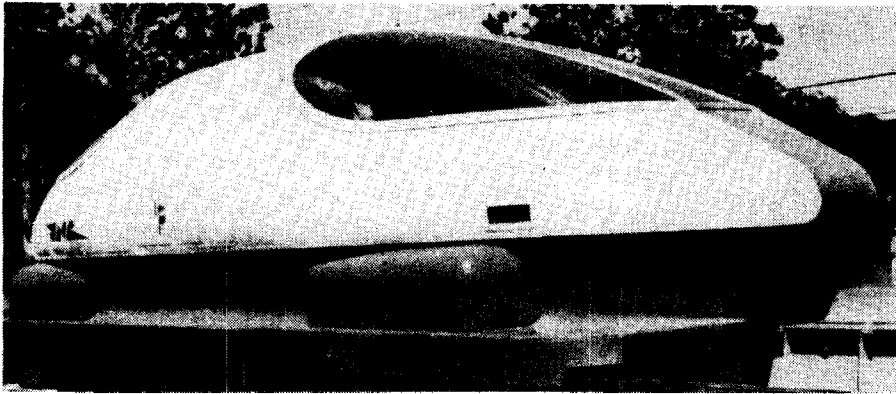


Fig. 9 Maglev vehicles can provide great speeds but they will require much more development and testing before they replace current forms of railway.

could move at speeds double those of today. Ship speeds are decided by physical limitations arising because of the wetted area surface drag and the disturbance caused to the water physically displaced by the ship's motion. Raising speeds above 10 knots or so demands enormous increases in fuel consumption. Solutions to this are to go completely under like the submarine for this also reduces the power needed, or rise out of the water on skis or the hover principle. One advantage of the latter is that the swell and roll of sea travel is greatly reduced. Hover ships can run today at speeds of 80 knots but problems with skirt sealing in rough weather appears to be a major current limit on usefulness on the open sea. The history of naval architecture, however,

shows that ships change in only minor ways and at a slow pace.

Finally, in the air: what will happen there? There was once a time — just a few years ago — when air travel had become the pinnacle of transport comfort and speed. But today it fails to provide a fast enough overall journey time because of airport regulations of arrival before departure time, clearance at security barriers, checking of tickets, settling of passengers and the like. It seems, as a rule of thumb, that the actual flight time is about equal to the sundry time involved for short international flights. Plans were published in the early seventies of cabins (in which people were assembled) ready to be attached to the plane. In some countries super highways link the remote airports

from the city centres. Somehow these plans did not provide the answer now seen to be needed.

Supersonic transportation is finding difficult acceptance, and it is just not possible to state at present whether it is good or bad with any degree of certainty. Only time will tell; hypotheses need verification by experiment.

The air ship is a strong contender especially if a design needing less helium were invented (certainly not one based on hydrogen, for it is highly inflammable when oxygen is available). The hot air balloon is thirsty for energy due to huge heat losses from the balloon surface — new materials and processes may provide us with an insulated enclosure of light enough structure for these to become viable for long flights.

Several companies have been formed to exploit commercial freight transport by airships. The idea is appealing — quiet, safe, speedy, not plagued with terrain problems and capable of loads equalling many semi-trailer cargoes put together. Airships are an example of past design hopes being reborn due to better technological availability.

In the next part we will look at communications and entertainment developments of the future.

FUTURE READING

We have recently learned from our New York office that a book, "People's Almanac" by Doubleday, New York has just been released in the US. This giant work looks into numerous aspects of the years ahead.

(To be continued...)



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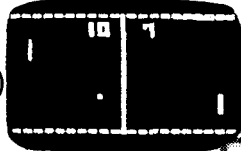
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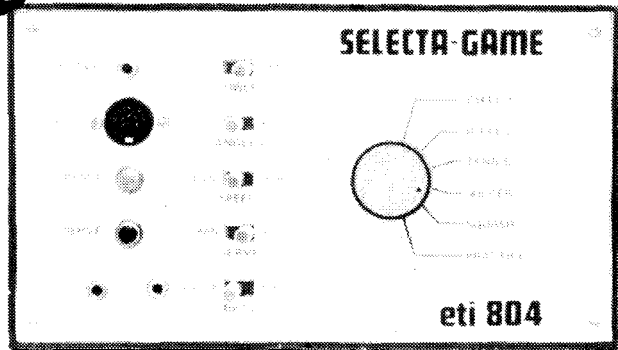
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This low-cost yet sophisticated TV game contains just one main IC plus a handful of other components yet out-performs virtually all other units currently on the market.

IT IS ABOUT A YEAR NOW SINCE TV games first appeared in Australia. Initially these units retailed for around a hundred dollars and had fairly limited capabilities. Many of our readers requested a TV game project but our investigation showed that 20 to 30 CMOS ICs would be required. As the circuit is quite complex we felt that the chances of a hobbyist building such a unit without problems were small, and any problems encountered would have been likely to be beyond solution with the equipment and knowledge available to the average constructor. We therefore decided not to do the project until single-chip TV game ICs became available. We knew that these chips were being developed and that they would make the project much simpler from the constructional point of view.

This project is based on such a single-chip device type AY-3-8500 from the General Instrument Corporation. The chip offers a choice of six games together with on-screen scoring and sound effects. The games are tennis, soccer, squash, practice and two rifle games. The rifle games require a 'rifle' which has additional circuitry built into it. If there is sufficient demand we will describe the construction of such a rifle at a later date.

Some additional circuitry, including two extra ICs, is required to build the game but the complexity of the complete circuit is still greatly reduced by the use of this particular IC. In

addition, the chip, although expensive, does allow the cost of the unit to be reduced considerably even though its performance is superior to many other games on the market.

Construction

The TV game employs some VHF circuitry which demands correct layout if proper operation is to be obtained. For this reason the game should only be built onto the printed-circuit board specified.

Commence construction by installing the seven tinned-copper wire links and then the low-height components (resistors, diodes, etc). Next install the capacitors and the transistors. ICs 2 and 3 are CMOS devices and should only be removed from their protective packing when you are ready to install them. Handle them as little as possible and when inserted solder the power supply pins (7 and 14) first. The main IC is expensive and it is therefore recommended (but not essential) that a 28-pin IC socket be used to mount it.

The coils L2 and L3 should now be constructed as detailed in Table 1 and then soldered into position making sure that L2 is oriented correctly.

The rotary switch may now be mounted in the following manner: First solder 25 mm lengths of tinned-copper to each of the switch pins (14 in all). Now orientate the switch correctly and feed the wires through the respective holes in the printed-circuit board, press

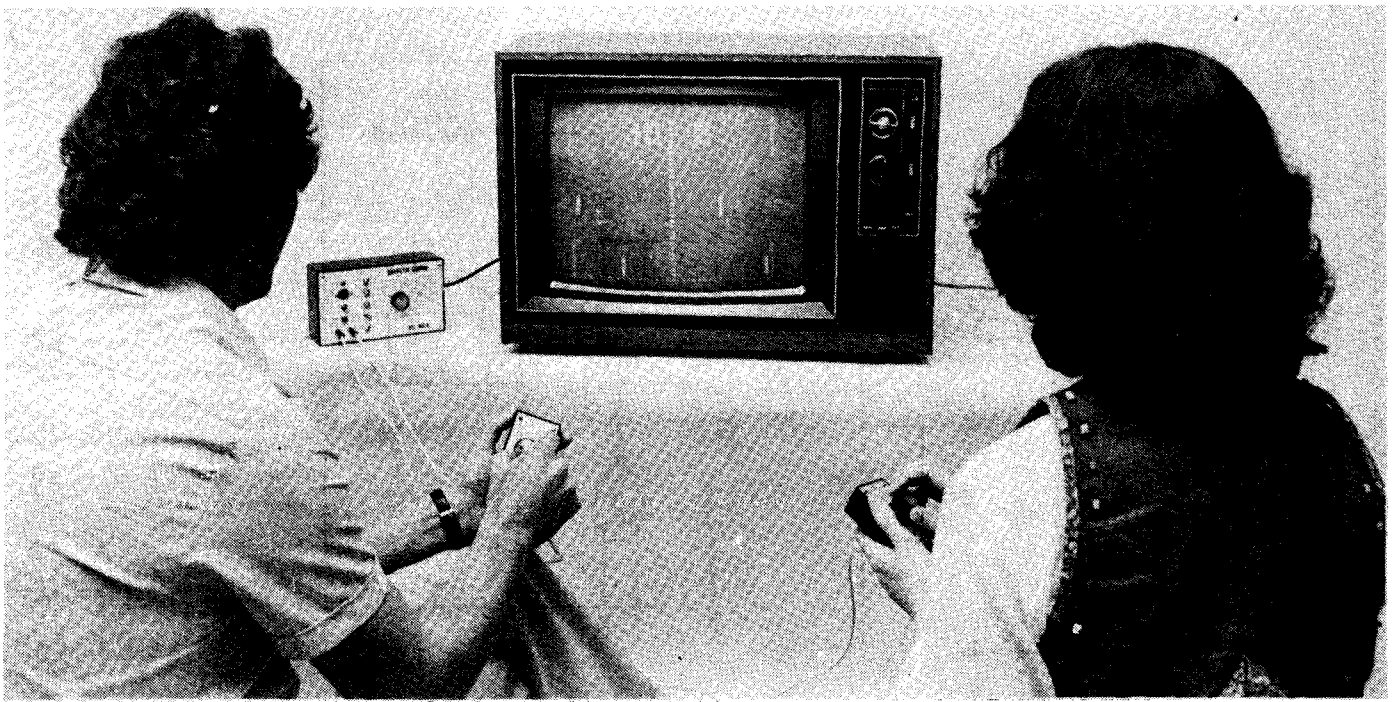
the switch down onto the board and solder all the wires to the tracks of the board.

Now prepare the push buttons, the 5-pin DIN socket and the phono socket by soldering 40 mm lengths of tinned-copper to each of the terminals. Feed the wires through the respective holes in the printed-circuit board but do not solder just yet.

The slide switches should also be prepared in the following manner: Cut 60 mm lengths of 20 gauge BS tinned-copper wire (largest gauge that will fit through the switch holes) and thread them through the holes in the switch pins so that pairs of poles are linked together. Centre the wires in the lugs and then solder them to the lugs. Now bend the wires down on either side and insert them in the holes provided in the printed-circuit board but do not solder at this stage.

Fit the front panel to the rotary switch (use a spacer washer) making sure that the board is square to the front panel and that there is enough clearance for the RF coil and the shield which have yet to be fitted. Attach the phono socket, the DIN socket and the push buttons to the front panel and then solder their leads to the board. Push the slide switches up against the front panel, line the switches up with the openings in the front panel and, making sure that the switch doesn't move, solder the leads to the board.

Now remove the front panel and



fit the 75-ohm output coax and the coax for the bats to the printed-circuit board. Feed the bat cables through rubber grommets in the front panel after first tying knots in them to prevent them being pulled through accidentally. Alternatively 3.5 mm jacks may be installed on the front panel for the bat outputs and the cables fitted with plugs so that they can be unplugged when the game is not being used.

Add the battery leads and connect the speaker by means of 150-mm long wires. Check all wiring and solder joints before fitting the main IC to its socket.

Before the shields for the RF stage

are fitted the unit should be connected to a TV set and aligned and checked as detailed in the alignment section.

After alignment is satisfactorily completed fit the component-side shield using four short lengths of tinned-copper wire and then fit the copper-side shield by simply soldering it to the copper earth plane in four or five places. Make sure that the shields do not touch any other tracks, leads or components which would cause a short.

The alignment of the unit may now be peaked up if required. The front panel is normally at +7.5V due to the connections to the phono sockets and

the shield is at 0V. Some plastic insulation tape should be used over the top of the shield or on the front panel to prevent shorting. Now fit the front panel and mount the assembly in the box. The batteries and speaker should be fitted into the bottom of the box under the printed-circuit board. Holes should be drilled in the box under the speaker to act as a grill to let out the sound from the speaker.

We initially used four "C" size batteries for power as the unit will work from about 5 to 8 volts. However to increase battery life 5 cells of either "C" or "AA" size should preferably be used.

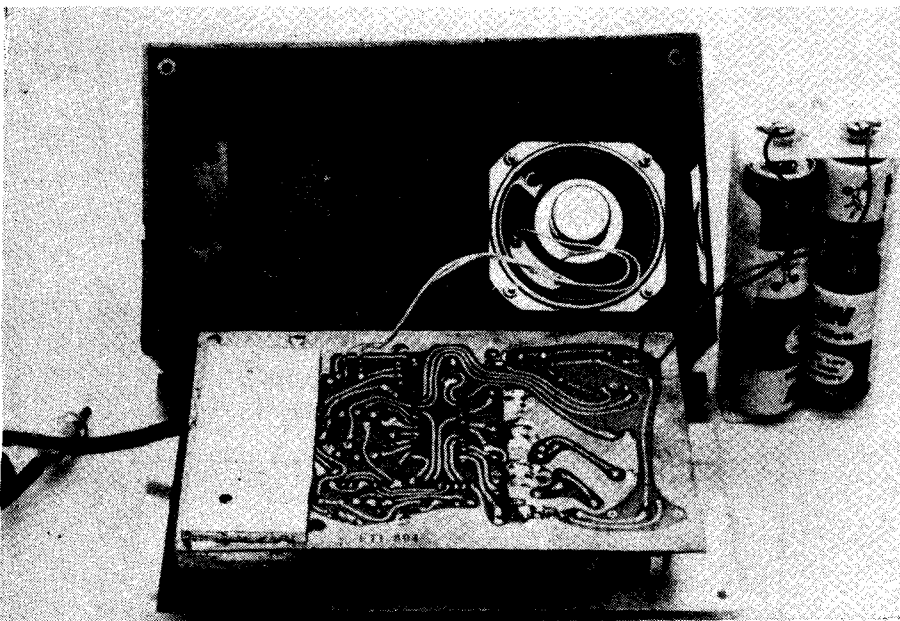
If an external power unit is used either 6 or 7.5 volts dc will operate the unit and the 3.5 mm phone jack used should be used with the +Ve lead on the common terminal.

Alignment

Switch the TV set to channel 6 (5 or 7 could alternatively be used if channel 6 is used in your area), connect the TV game to the antenna input of the set and switch both units on.

Press the reset button on the game and tune coil L2 until the set appears to be receiving the signal. (At this time the picture may appear to be just a series of dots). Adjust the trimmer capacitor CV1 until the picture locks. Then it may be necessary to readjust L2 for the best picture.

When performing these adjustments it is best to use non-metallic tools so that the tuning point does not alter when the tool is removed.



SELECTA-GAME

ELECTRONICS TODAY INTERNATIONAL - NOVEMBER 1976

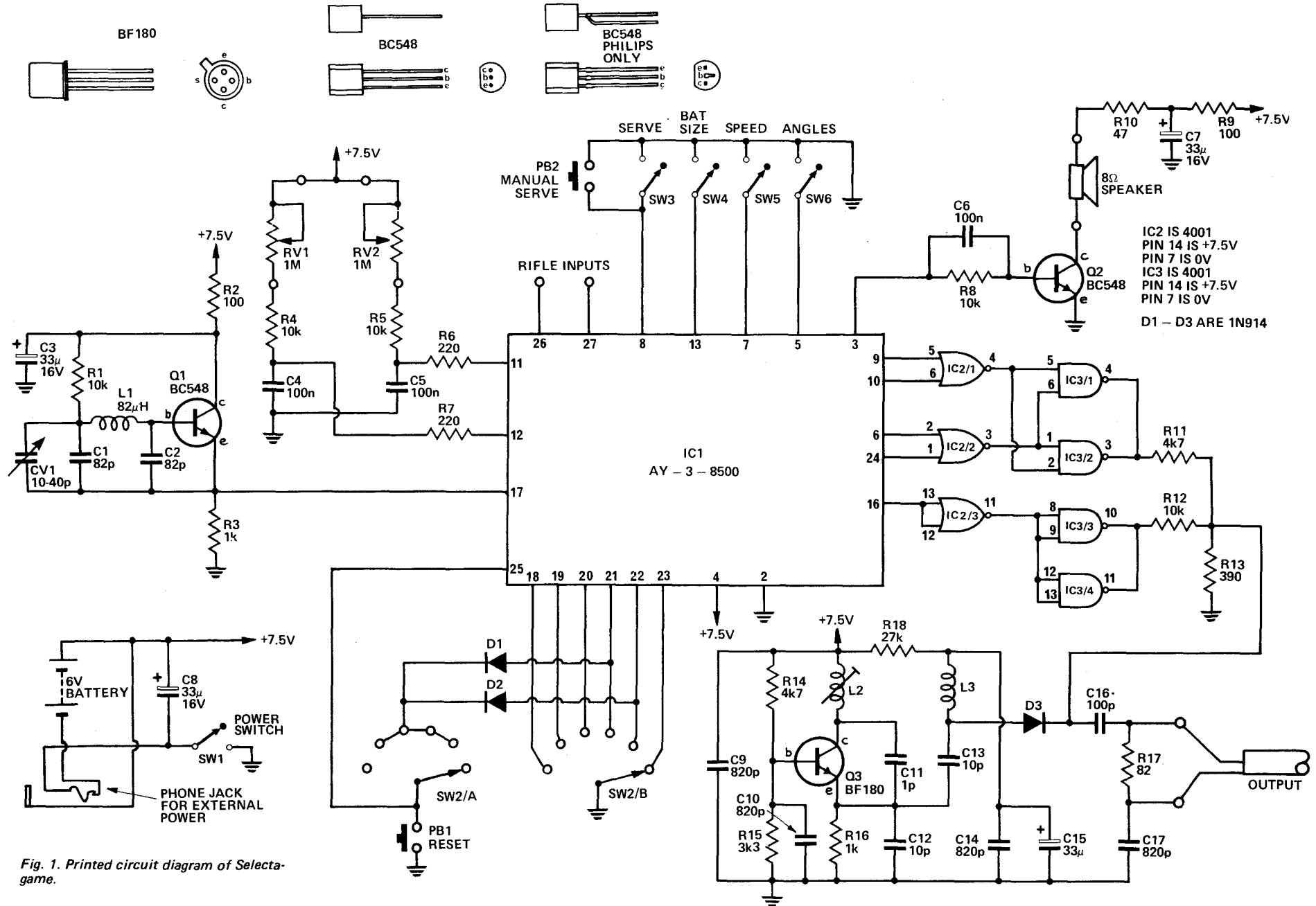


Fig. 1. Printed circuit diagram of Selecta-game.

How It Works — ETI 804.

Unfortunately the manufacturers don't give much information on how the main IC works — we are only told how to use it. The chip is obviously a digital IC (because there are two ball speeds, the rebound angles are defined and there is no provision for variable speed or bounce).

A 2 MHz oscillator is required for the chip to derive the synchronising pulses required for line and frame synchronisation of the TV set. This oscillator is provided by Q1 and its associated components with CV1 providing calibration.

The bats are simply one megohm potentiometers connected as variable resistors which effectively vary the charging time of capacitors C4 and C5. The capacitors C4 and C5 are discharged by the chip at each frame sync pulse and the time taken to charge again (as set by the bat pot setting) determines the vertical position of the bats on the screen. The bat size, ball speed, deflection angles and serve are simply selected by connecting the appropriate pin of the IC to '0' volts.

Outputs from the chip are left and

right bat, sync, ball, score and sound — all on separate pins. The bats, ball and score outputs are combined by IC2/1,2 and IC3/1,2 to produce a composite video signal. The sync pulse is buffered by IC2/3 and IC3/3,4. The sync and information pulses are then added by R11, 12 and 13. The sound output is buffered by Q2 to provide the power necessary to drive the speaker.

So that the game may be fed into the antenna terminals of a TV receiver the video signal must be modulated onto an RF oscillator tuned to the

desired channel (176 MHz for channel 6). Transistor Q3 and its associated components form the required oscillator. The oscillator is then modulated with the composite video by means of the diode modulator D3.

The oscillator and the modulator are screened by means of shields to prevent the RF from causing interference to other TV sets (and to prevent other TV sets from interfering with the game). These shields also minimise detuning effects when the hand is brought close to the oscillator.

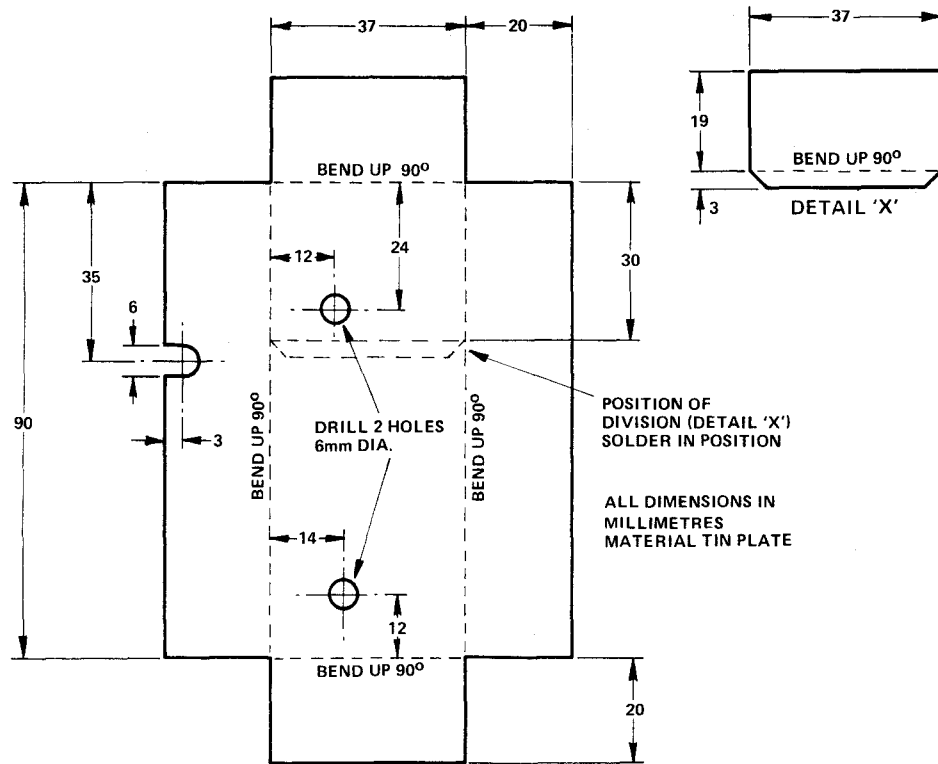


Fig. 2. Dimensions of shield on component side.

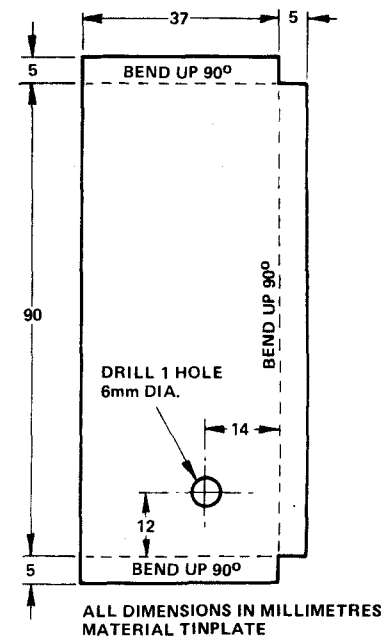
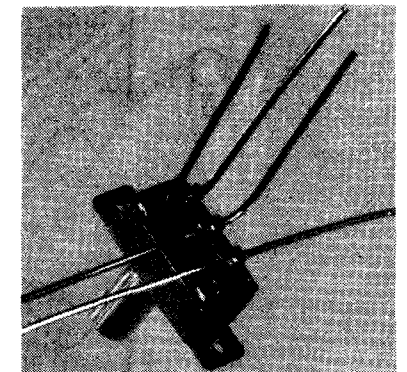
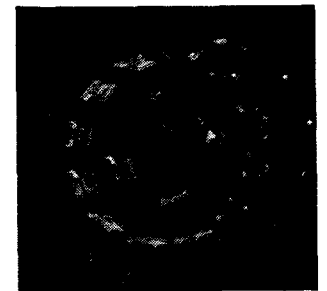


Fig. 3. Dimensions of shield on copper side.



Photographs showing wires attached to switches before installation.

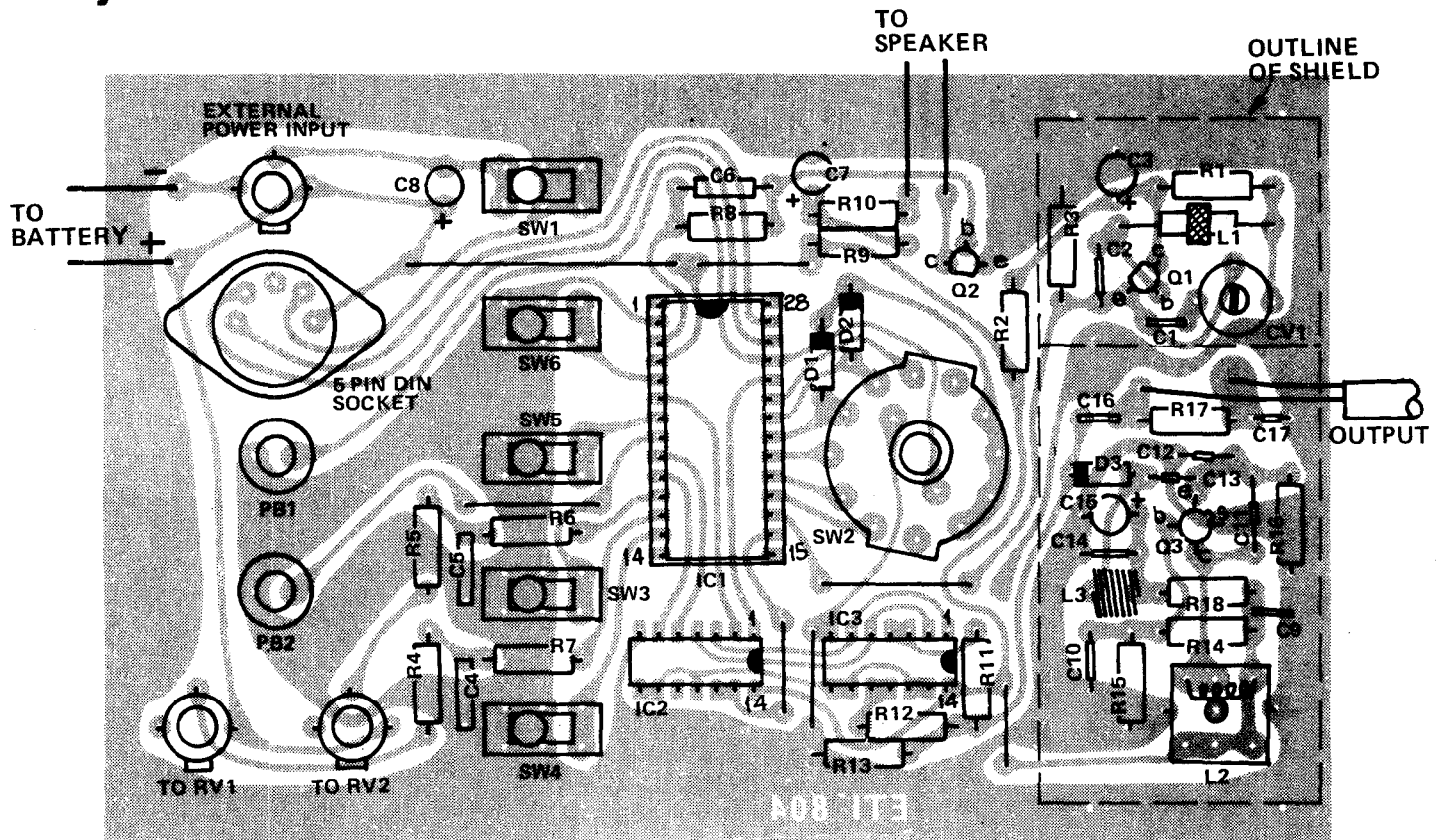


Fig. 4 Component overlay of Selecta-game.

PARTS LIST ETI 804

Resistors all 1/4w 5%

R1	10 k
R2	100
R3	1 k
R4,5	10 k
R6,7	220
R8	10 k
R9	100
R10	47
R11	4 k 7
R12	10 k

R13	390
R14	4 k 7
R15	3 k 3
R16	1 k
R17	82
R18	27k

Potentiometers

RV1,2	1M lin rotary
-------	---------------

Capacitors

C1,2	82 p ceramic
C3	33 μ 16 v electro
C4,5,6	100 n polyester
C7,8	33 μ 16 v electro
C9,10	820 p ceramic
C11	1p0 ceramic
C12,13	10 p ceramic
C14	820 p ceramic
C15	33 μ 16 v electro
C16	100 p ceramic
C17	820 p ceramic

Variable capacitor

CV1	10-40p
-----	--------

Transistors

Q1,2	BC548 or similar
Q3	BF180

Diodes

D1-D3	1N914
-------	-------

Integrated Circuits

IC1	AY-3-8500
IC2	4001 (CMOS)
IC3	4011 (CMOS)

Inductors

L1	82 μ H RF choke
L2	See table 1
L3	See table 1

Miscellaneous

- PC board eti 804
- 2 pole 6 position switch
- Five slide switches
- 8 ohm speaker
- 3.5 mm phone socket
- 5 pin DIN socket
- Two miniature push buttons
- Three knobs
- One large box 196mm x 113mm
- Two small boxes 83mm x 54mm
- Single "C" size battery holder der
- 28 pin IC socket

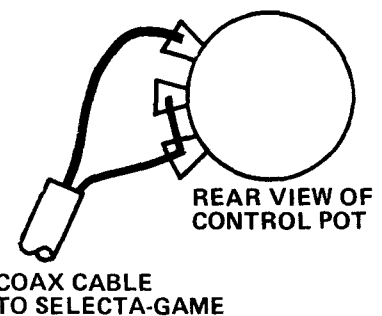


Fig. 5. Diagram showing wiring of control pot.

TABLE 1 ETI 804

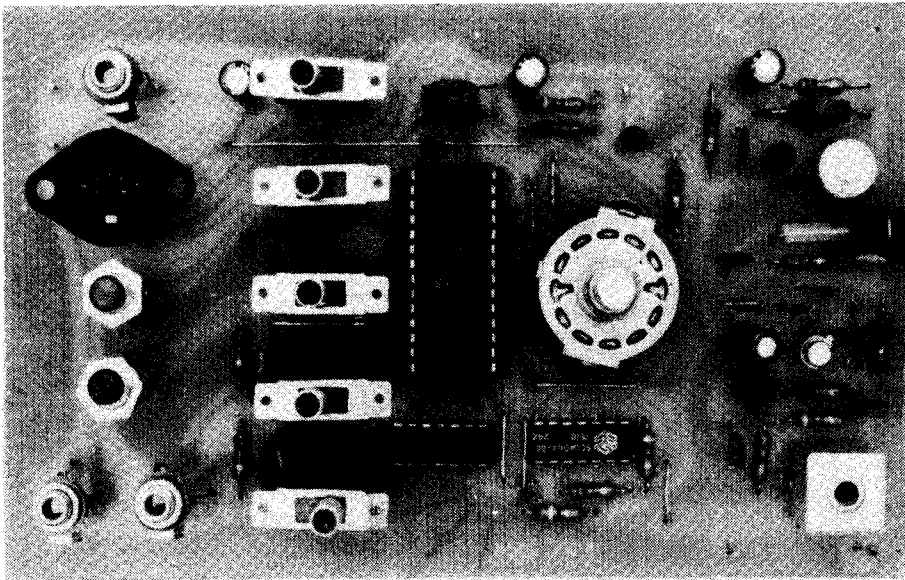
Winding details of coils L2 & L3

L2	Former 5mm	Neosid 722/i
	6 pin base	Neosid 5027/6 PLB
	Can	Neosid 7100
	Slug	Neosid 4 x 0.5 x 10 F29

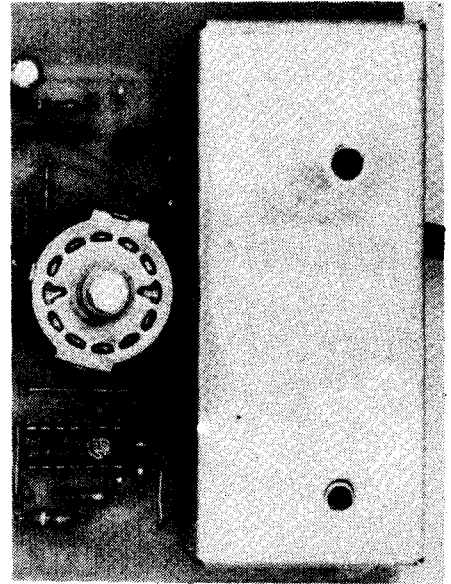
Winding 4 turns close wound 24 B&S

L3

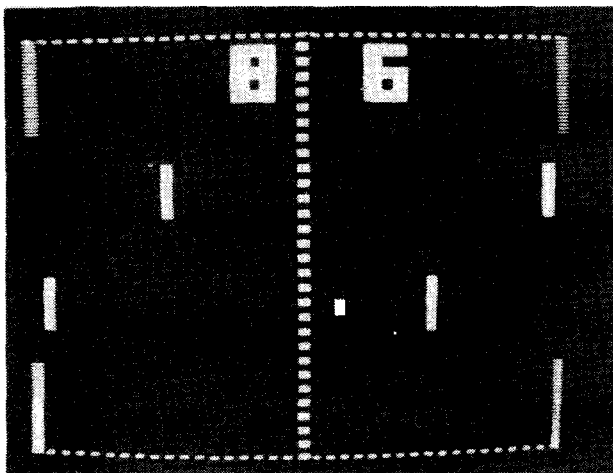
6 turns 24 B&S wire close wound about 5mm diameter, air core. (wind on a former, ie a knitting needle or drill, then remove former)



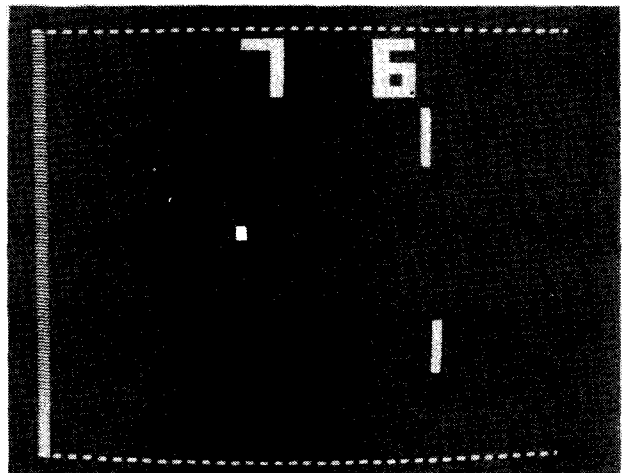
Photograph of completed board less shield.



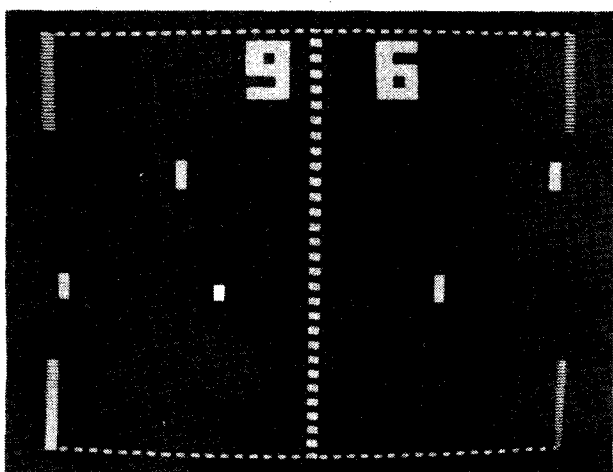
Photograph showing shield fitted. Note two adjustment holes



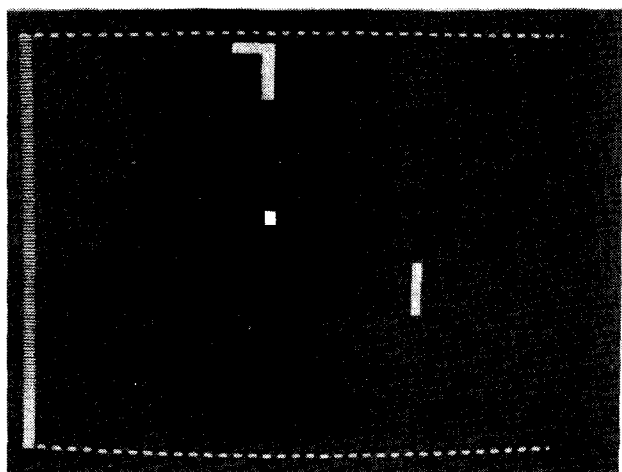
Soccer



Squash



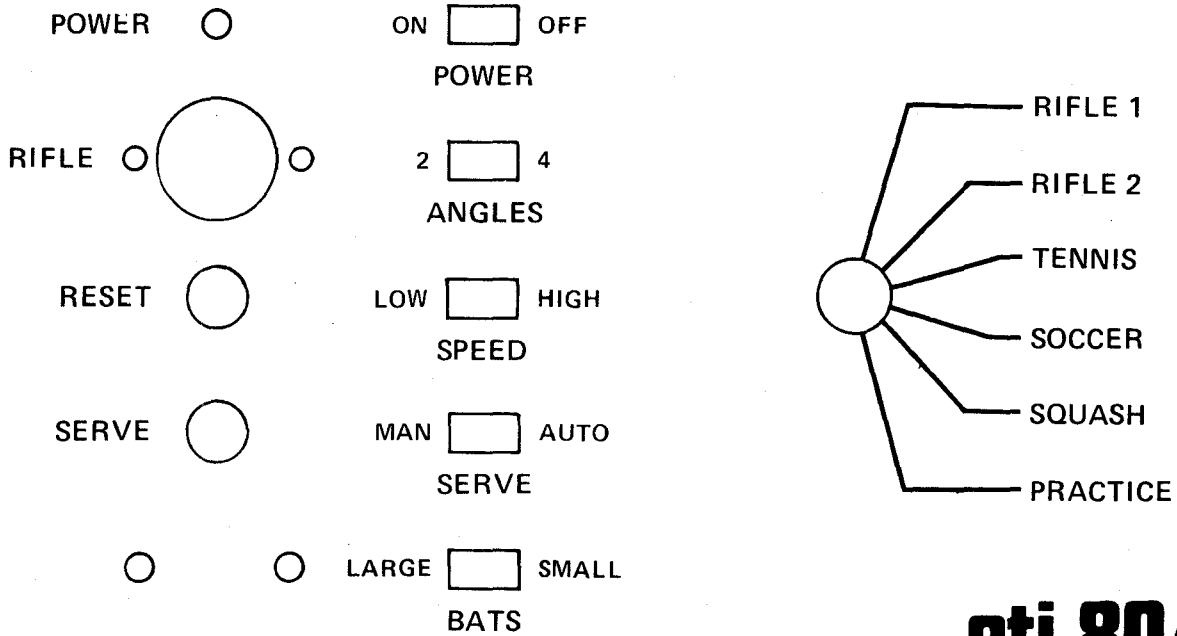
Soccer with small bats



Practice

These photographs show some of the games that can be played.

SELECTA-GAME



eti 804

Fig. 6. Front panel layout. Full size 190mm x 107mm.

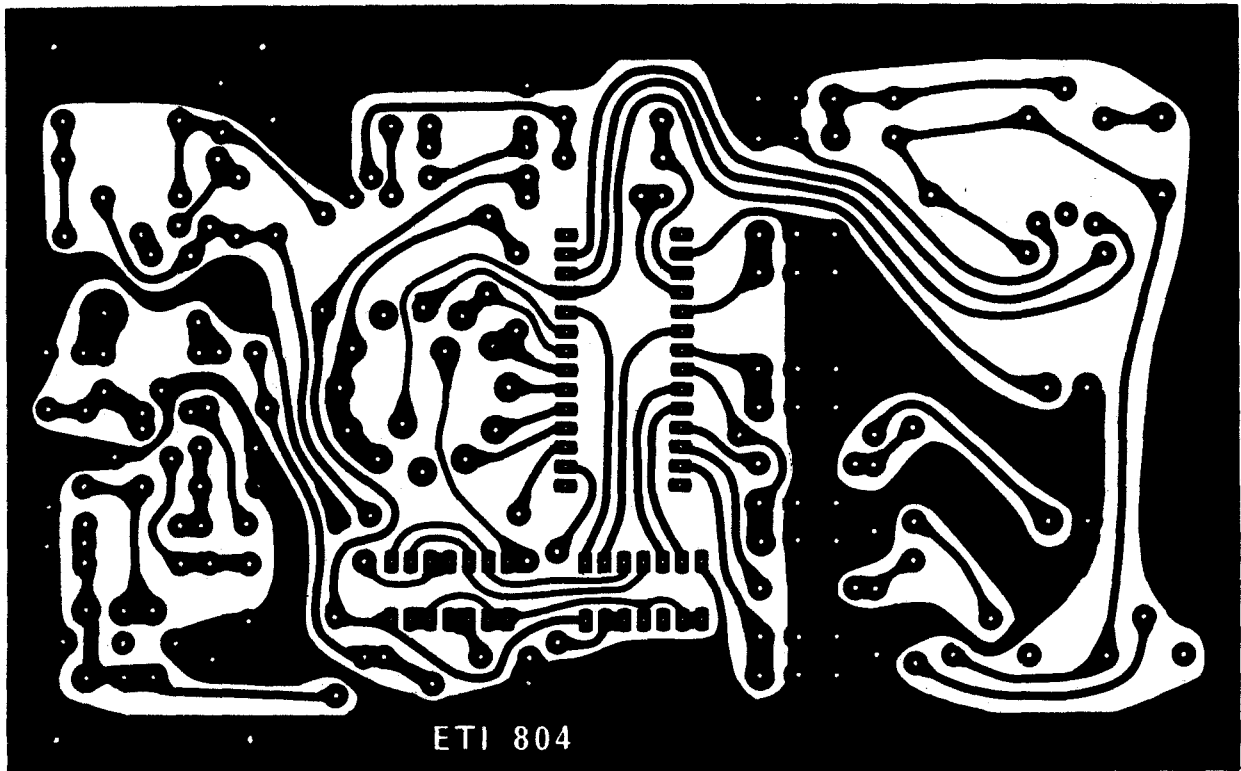


Fig. 7. Printed-circuit layout. Full size 163mm x 102mm.

THE GAMES

1) Practice: The ball reflects off the end wall and the side walls — the player has to stop it exiting the screen on the right-hand side. This game is an electronic version of hand-ball or squash with only one player.

2) Squash: This game is like the practice game but now there are two players who take turns in hitting the ball. The bat has no effect on the ball when it's your opponent's turn.

3) Soccer: The ball reflects off all four sides of the pitch, except for the goal mouth. The goalkeeper defends this in

the same way that the bats defend their court in the other games. The player has a second man on the field for soccer — the forward. This man will act like a bat when faced with a ball moving towards his own goal — the ball reflects towards his opponent's goal — but when the ball is moving in the right direction (towards his opponents side) the ball passes through him. However by careful positioning of the forward the ball can be made to deflect towards the goal.

4) Tennis: The game of TV Tennis is widely known and on Selecta-game the only unusual features are those listed in the specification.

SPECIFICATIONS — ETI 804

Output	Picture: TV signal on channel 6 (can be set up on any other channel). Sound: Three audio tones indicate hit, bounce and score. Reproduced from a loudspeaker in Selecta-Game.
Players' Controls	Each player uses a single rotary control to position his bat/men on the screen. In the practice game one control operates; for tennis, soccer and squash two players each have a control. For the rifle games a special rifle is needed (not described in this article).
Game Selection	Basic Games: 1) Practice 2) Squash 3) Soccer 4) Tennis Other Games (these cannot be played without a special rifle): 5) Rifle—1 6) Rifle—2
Scoring	On-screen scoring up to a maximum of 15 points.
Other Features	Two ball speeds Two bat sizes Two angles $\pm 20^\circ$; or four angles $\pm 20^\circ$ & $\pm 40^\circ$. Manual or automatic service

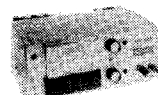
USING YOUR SKILL

With the 'Angles' switch at '2' the ball moves across the screen at $\pm 20^\circ$ from the horizontal. When hitting the sides and walls of the court the laws of reflection are obeyed. When the ball hits the bat this isn't always the case: a ball hitting the top half of the bat will leave with an upward trajectory, and a ball hitting the bottom half of the bat will bounce downwards. This effect can be utilised by the skilful player —

when the soccer forward is used to change the direction of the ball as it approaches the goal it usually beats the goalkeeper.

With the 'angles' switch at '4' the game becomes even more exciting. Now the bat has to be divided into quarters: starting from the top and working down the angles of the emerging ball are $+40^\circ$, $+20^\circ$, -20° , -40° . And if you can cope with that try switching to small bats and high speed ball!

ETI 131 Power Supply (See April '76 ETI)

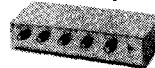


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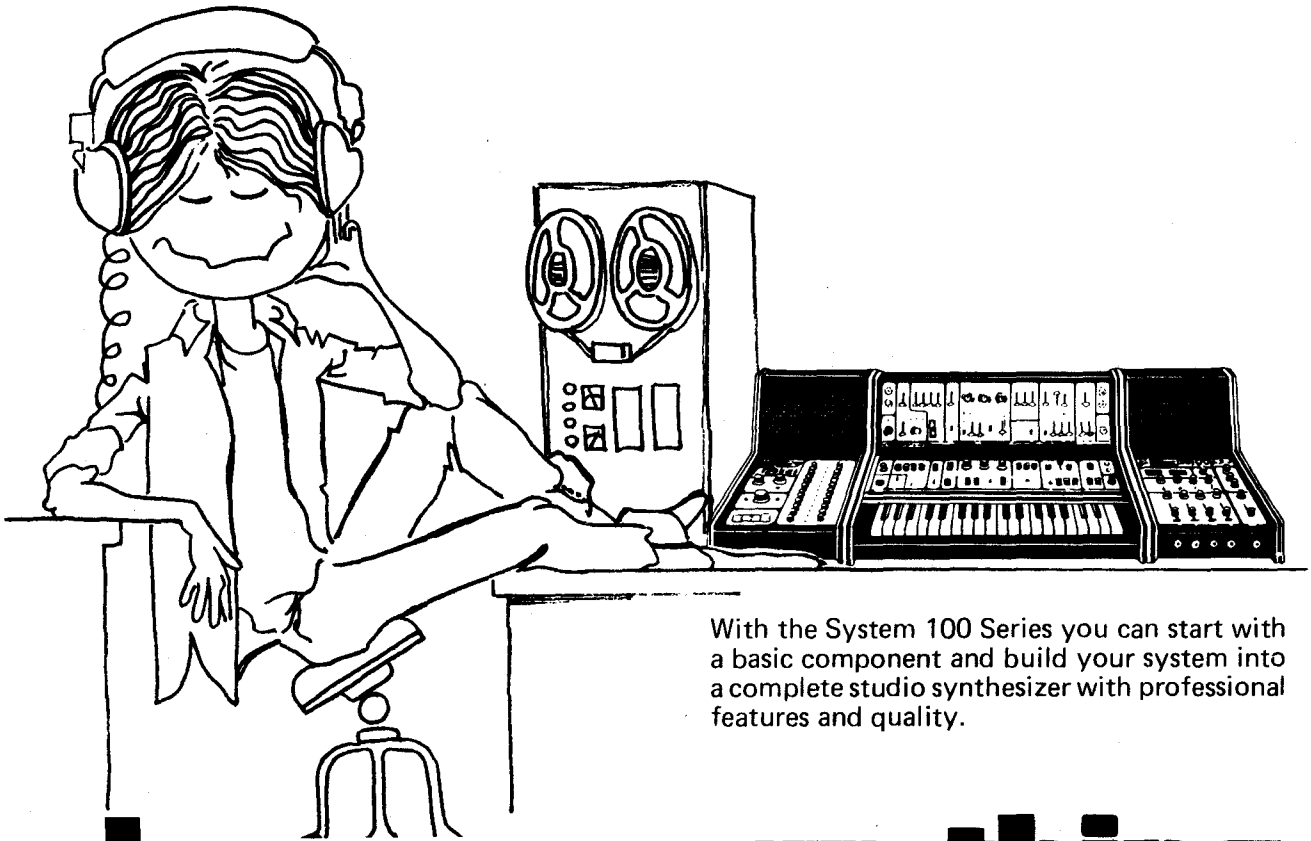
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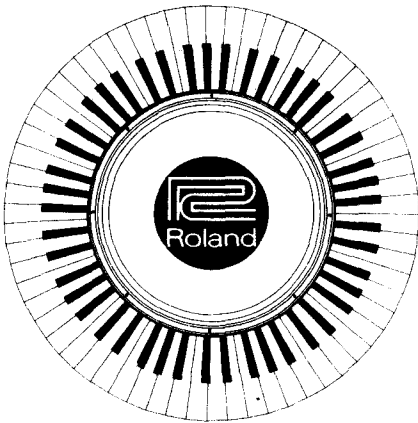
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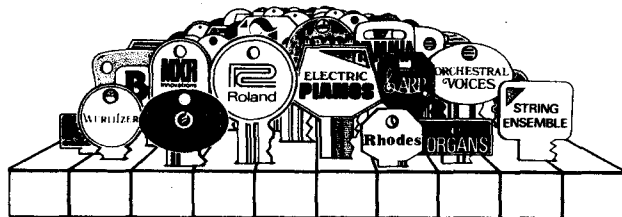


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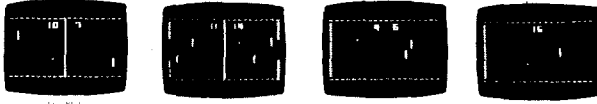
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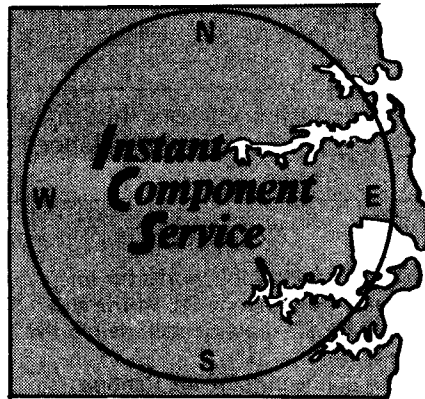
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Construction

A pc board is not necessary but it does make the assembly much easier, and it reduces the risk of mistakes. Assemble the components to the board according to the overlay in Fig. 3. Watch the orientation of the BC 549 transistors as there are two different pin-outs depending on the manufacturer. The Philips is shown on the overlay. Note also that the IC is marked on one end and this should be orientated as shown.

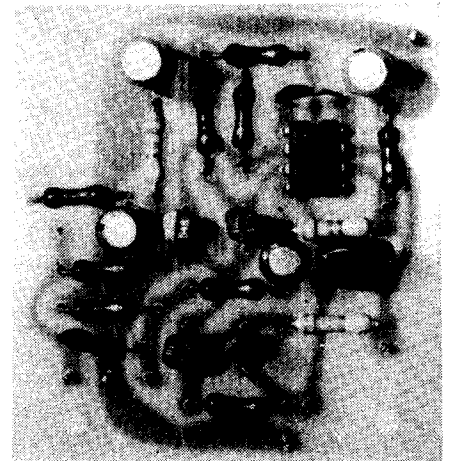
As the unit is to be used with low level inputs it will be necessary to use shielded cables on the input and it may also be necessary to shield the complete unit to eliminate unwanted pick up from the mains.

Although the unit is designed for 600 ohm input and 40 dB gain (i.e., a gain of 100) other input impedances and gains can be handled —
 $R1 = R4 = \text{input impedance divided by two.}$
 $R5 = R13 = \text{voltage gain times the value of } R3.$

The first equation works for impedances up to about 5k. Above this value $R2 + R3$ must be included in the calculation.

The connection of the Cannon socket is shown in Table 1.

The Common mode rejection ratio is determined by the accuracy of $R1, R4, R5$ and $R13$. If high rejection is needed this can be done in either of two ways: The resistors can be selected to be the same value ($R1 = R4, R5 = R13$) or 1% resistors can be used.



SPECIFICATION* ETI 449

Frequency Response	10 Hz – 20 kHz (<5 V output)	+0 dB –3 dB
	10 Hz – 10 kHz (10 V output)	+0 dB –3 dB
Gain	40 dB	
Equivalent Input Noise	–123 dB (0.5 μ V)	
Distortion	0.05% 300 mV – 5 V output 100 Hz – 10 kHz	
Max Input Voltage	100 mV	
Common Mode Rejection Ratio	60 dB	
Maximum Common Mode Signal	3 V	
Supply Voltage	+8 to +16 V	

*Measured performance of prototype with supply rails of ± 15 V – the figures should be typical.

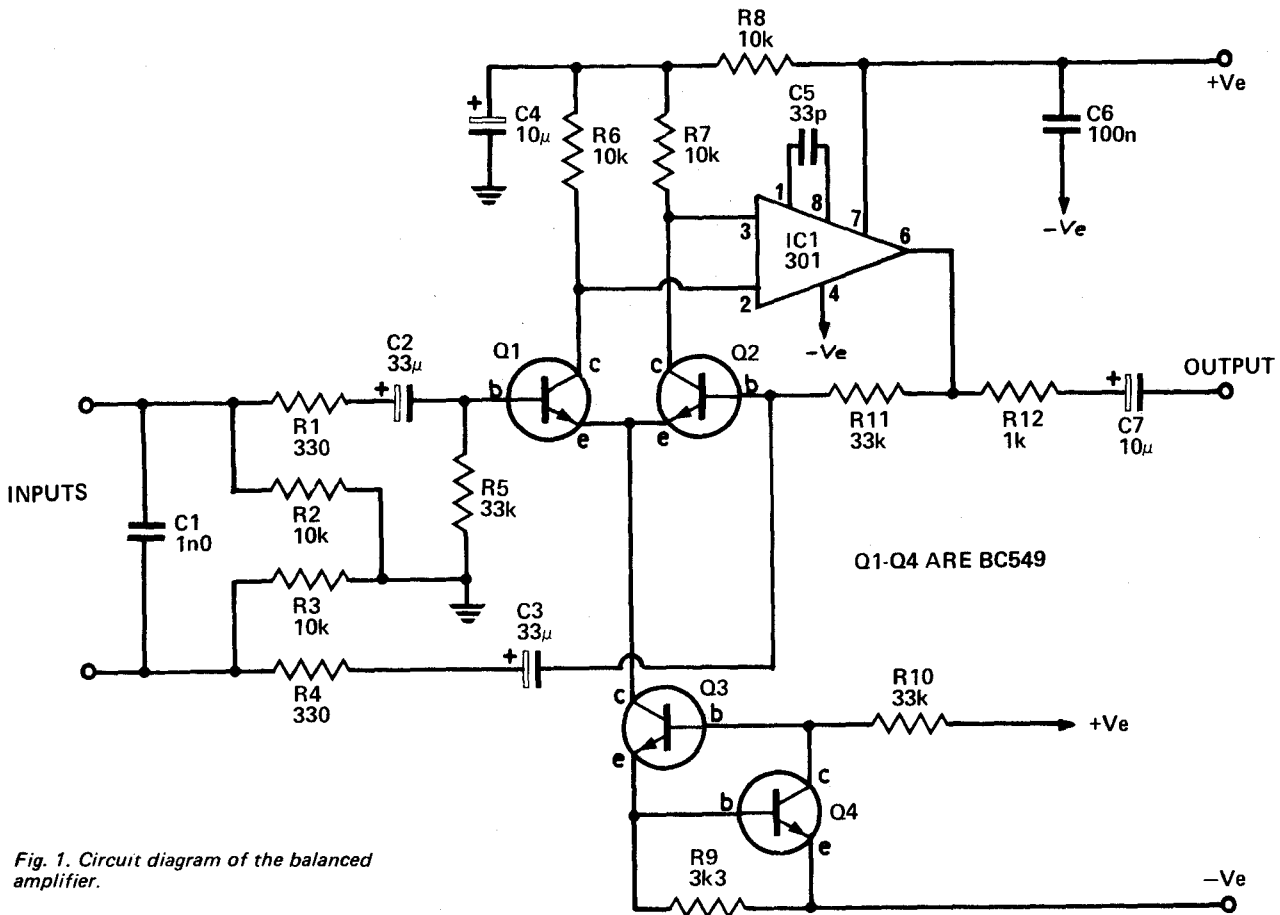


Fig. 1. Circuit diagram of the balanced amplifier.

HOW IT WORKS ETI 449

A "balanced" amplifier or differential amplifier has two separate inputs and only the difference between these inputs is amplified. To explain how this works refer to Fig. 2, which is a simplified version of the actual circuit. To make the maths easier we will reduce the gain to nine by making $R1 = R4 = 1$ and $R5 = R13 = 9$. The actual units are not important, only the ratio.

We will start the explanation by looking at the case where point B is at 0V and A is at +100 mV. An ideal amplifier does two things — it does not take any current into the input terminals and it adjusts the output to maintain no voltage difference between the input terminals. We therefore must have 100 mV across R4 and consequently a voltage of 900 mV across R13 (it has 9 times the resistance and the same current as R4). This gives a gain of nine. The output is therefore -900 mV.

In the case when point A is at 0V and point B is at +100 mV, point D

will be at $(V_B \times \frac{R5}{R1 + R9}) = 90 \text{ mV}$

Therefore point C will also be at +90 mV. The voltage across R4 will be 90 mV and voltage across R13 will be 810 mV ($9 \times 90 \text{ mV}$)

This means the output voltage must be +900 mV. This is also a gain of nine. Notice, however, that the polarity (or phase) is different.

Now suppose both inputs are at, say, +1V, point D will be at +900 mV and so will point C. The voltage across R14 is 100 mV and R13 900 mV. This gives an output voltage of 0 V. The common signal is not amplified in any way. If, however, one input (B) is at 1 V and the other (A) is at 1.01 V the difference is amplified and the output will be -1 V.

Getting back to the actual circuit, we have used an LM301A with two low-noise transistors in the front stage. These transistors are supplied with a constant current by Q3 and Q4. A constant current is needed as this allows the inputs to move up and down without changing the voltage across R6 or R7.

We have decoupled the supply to Q1 and Q2 by C4 to improve supply rail rejection.

The input resistors are decoupled by C2 and C3 to minimise the offset voltage of Q1 and Q2 being amplified by 100 (which could be excessive if matched transistors are not used). The capacitor C1 is to ensure the inputs are terminated at high frequencies. This ensures stability which would otherwise be a problem when the microphone is unplugged.

The resistors R2 and R3 refer the inputs to 0 V but are high enough not to affect the operation in any way.

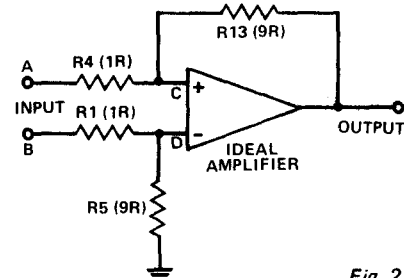


Fig. 2.

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Project 449

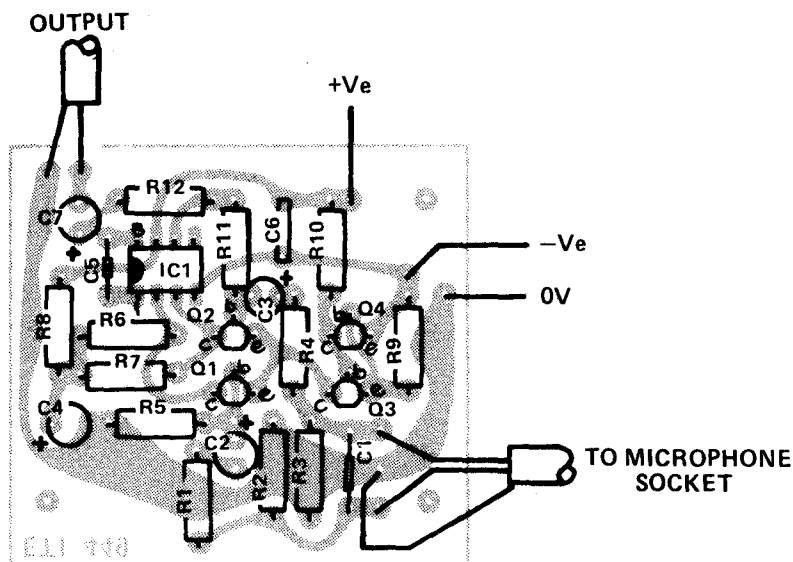


Fig. 3. Component overlay

PARTS LIST ETI 449

Resistors all 1 W 5%

R1	330
R2,3	10k
R4	330
R5	33k
R6,7,8	10k
R9	3k3
R10,11	33k
R12	1k

Capacitors

C1	1n0 polyester
C2,3	33μ 10v electro
C4	10μ 16v electro
C5	33p ceramic
C6	100n polyester
C7	10μ 16v electro

Q1-Q4	Transistors BC 549
IC1	LM301A

PC Board ETI 449

TABLE 1

Connection of Cannon plug for microphones

Pin 1	EARTH
Pin 2	BLACK INPUT connect to R1
Pin 3	RED INPUT connect to R4

FOR UNBALANCED INPUT CONNECT PIN 1 AND 2 TOGETHER ON MICROPHONE PLUG.

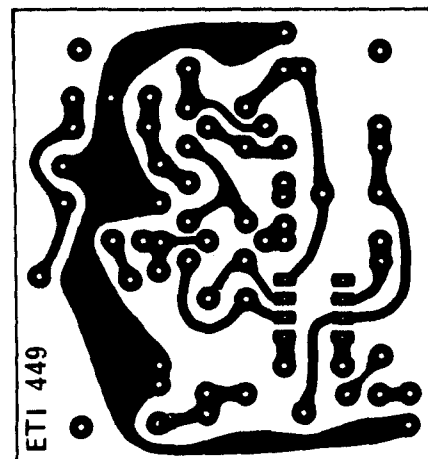


Fig. 4. Printed Circuit board layout. Full size 60 x 55 mm

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4025	Triple 3 input NAND	.30
4027	Dual J-K flip flop	.90
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7402	Quad 2 input positive NOR	.31
7403	Quad 2 input positive NOR with o/c outputs	.31
7404	Hex inverter	.31
7405	Hex inverter with o/c outputs	.31
7408	Quad 2 input positive AND	.31
7409	Quad 2 input positive AND with o/c outputs	.31
7410	Triple 3 input positive NAND	.31
7413	Dual NAND Schmitt trigger	.75
7420	Dual 4 input positive NAND	.31
7430	8 input positive NAND	.31
7437	Quad 2 input positive NAND buffer	.55
7440	Dual 4 input positive NAND buffer	.31
7441	BCD to decimal decoder/driver	1.10
7442	BCD to decimal decoder	.90
7447	BCD to 7 segment dec/driver with 15V outputs	1.25
7450	Expander dual 2 wide 2 input AND OR INV	.31
7451	Dual 2 wide 2 input AND OR INV	.31
7453	Exp. 4 wide 2 input AND OR INV	.31
7454	4 wide 2 input AND OR INV	.31
7460	Dual 4 input expander	.31
7470	Gated J-K flip flop	.45
7472	J-K master slave flip flop	.55
7473	Dual J-K master-slave flip flop	.80
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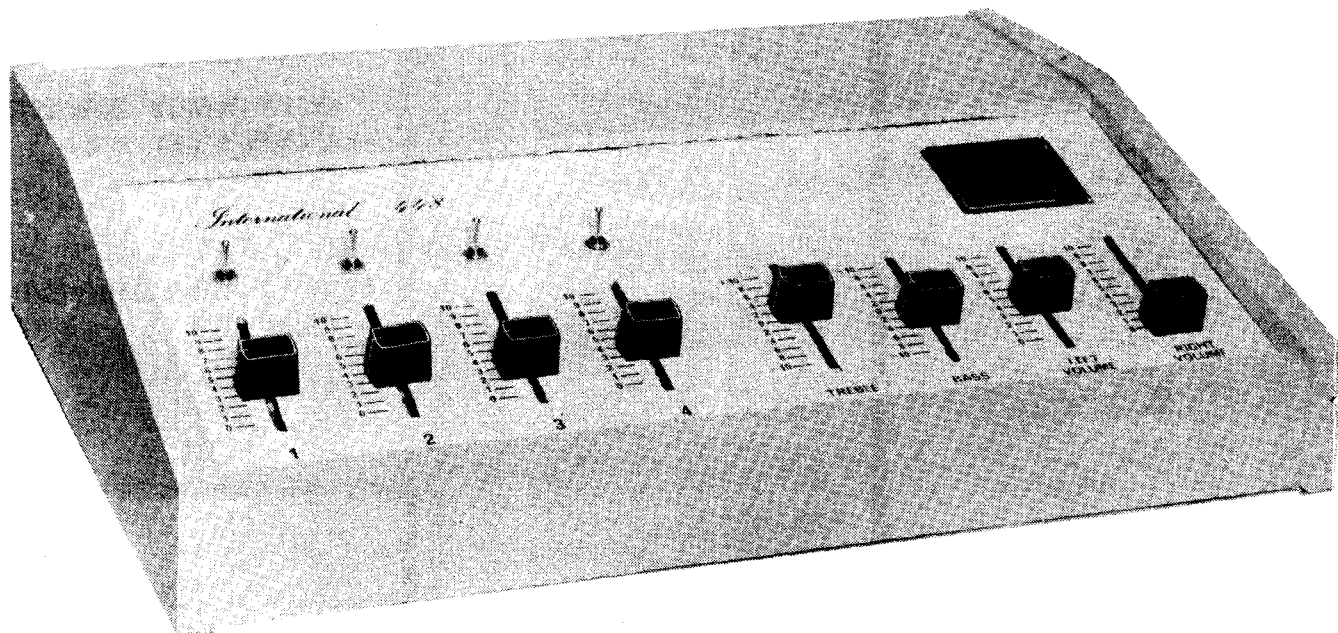
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DISCO MIXER

When we found out that more of our small mixer projects were being used in discos than anywhere else we decided it was about time we designed one specifically for the job.



THIS IS A GENERAL-PURPOSE mixer project which can be tailored by the constructor to meet the needs of a specific application. Some of the boards used have been published in previous issues of ETI; there are three new boards. The balanced-microphone preamp board is described elsewhere in this issue and in this article we introduce our disco mixer board (containing the power supply and stereo mixer amplifiers) and the ETI 448A headphone amplifier.

Using the boards listed below virtually any audio source can be mixed by the operator to provide a

stereo programme suitable for driving slave amplifiers (such as the ETI 413 100 W amps). The boards we suggest are:

ETI 448 — Power supply and main stereo mixer.

ETI 448A — Mono headphone amplifier.

ETI 449 — Balanced microphone preamplifier.

ETI 445 — General purpose stereo preamplifier (July 76 ETI).

ETI 446 — Audio limiter (in August 76 ETI).

The general purpose preamp can be configured to give a response

which is flat, which follows the RIAA phono curve, or which follows the NAB tape curve. This board can be used to take the output of magnetic cartridges on the record deck, the output of a cassette head, or the input from the DJ's microphone. If a high impedance microphone is used this can be connected directly into the input of ETI 446 limiter.

The output of a cassette machine, or any other high-level source can be connected directly to the ETI 448 mixer.

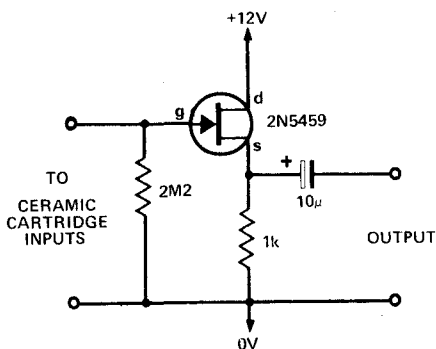
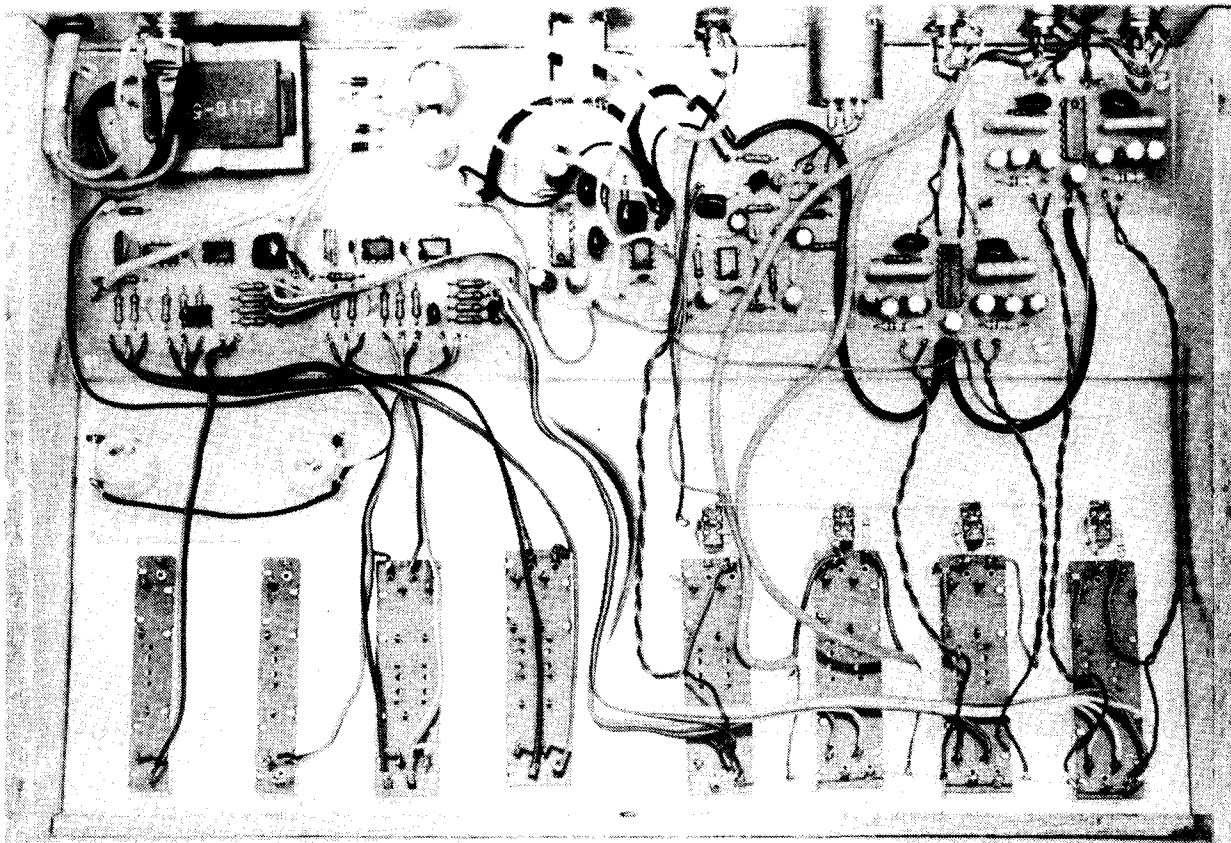


Fig. 1. Circuit diagram of ceramic cartridge preamp.

CONSTRUCTION

Before beginning construction, decide on how many inputs you need and what types. For our prototypes we chose four inputs but there is no need to limit the mixer to this size — the only consideration is the size of the box. Also decide what type of sockets you want, as this will affect the rear panel. We have not shown a drawing of the rear panel because there are too many possible ways of designing this.

Assemble the boards with the aid of the overlay drawings shown in this article and those in previous months (for the preamp boards). The photo shows the positioning of the pc boards, although this is not critical and the boards can be mounted on the top

TABLE 1

SUGGESTED PREAMPLIFIERS ETI 448

Phono	ETI 445 (RIAA)
Magnetic	see Fig. 1.
Ceramic	
Tape	None required
Tape Head	ETI 445 (NAB)
Microphone	
Low impedance	ETI 449
47 k	ETI 446 (R2 = 4 k7)
Microphone Limiting	
Low impedance	ETI 449 + ETI 446 (R2 = 15
47 k	ETI 446 (R2 = 47 k)

SPECIFICATION ETI 448

No. of inputs	Nominally 4
No. of outputs	2 main signal outputs 1 headphone amplifier output
Tone controls	Overall bass and treble
Output noise (Mixer stage only)	1 mV (mainly hum)
Maximum output voltage	6 V

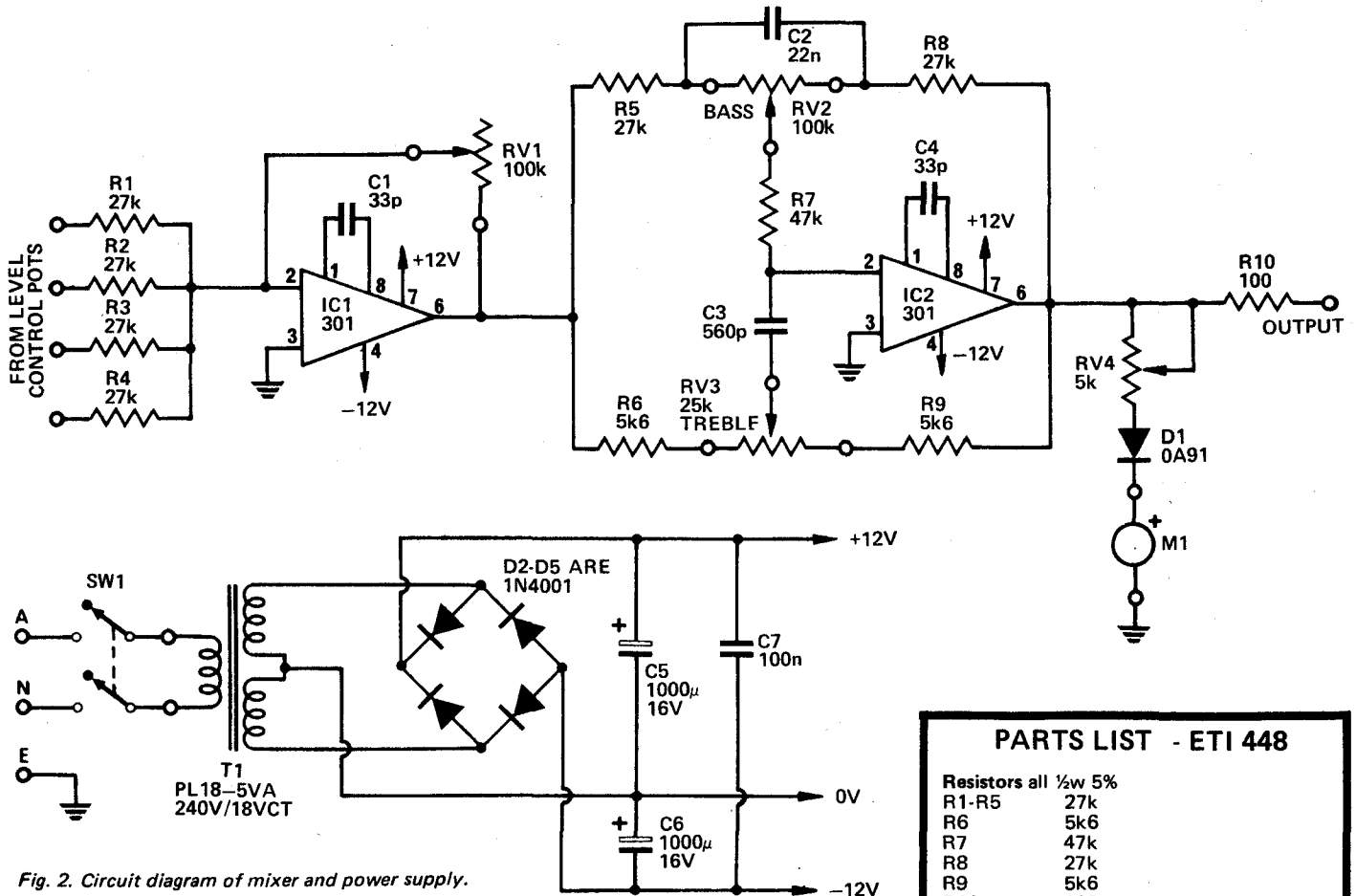


Fig. 2. Circuit diagram of mixer and power supply.

PARTS LIST - ETI 448

Resistors all 1/2w 5%
 R1-R5 27k
 R6 5k6
 R7 47k
 R8 27k
 R9 5k6
 R10 100

Potentiometers

RV1 100k log single gang slide 45mm
 RV4 5k trim

Capacitors

C1 33p ceramic
 C2 22n polyester
 C3 560p ceramic
 C4 33p ceramic

IC1, 2 LM301A
 D1 OA91

Two of all the above components are required for stereo operation.

RV2 100k lin dual gang 45mm slide
 RV3 25k lin dual gang 45mm slide
 RV5-RV8 10k log dual 45mm slide

C5, 6 100µ 16V electro
 C7 100n polyester

D2 - D5 1N4001 or similar

Transformer PL18-5 VA (18 V CT) (or similar)

pc board ETI 448

Switch 2 pole 2 position 240 V toggle

Dual VU meter

Four preamps to suit inputs used.
 Case to suit
 Sockets to suit

HOW IT WORKS

The inputs from the turntables, tape recorders, microphones, etc., must be amplified, and if necessary equalized, by a preamplifier before any of the controls can handle them. Our recommendations for suitable preamps are given in Table 1. The output of each of these preamps is adjustable, by means of a volume control or fader, before being mixed in IC1. The overall gain of the mixer stage is adjusted by means of RV1. If different preamps have widely differing output voltages the value of R1-R4 can be changed to make them match.

The output of IC1 goes then to the tone control stage, IC2, which normally has a unity gain when the controls are centered. However, this gain is adjustable, with respect to frequency, if the tone controls are not centered. The output of the tone control stage directly drives the main power amplifiers. This output

is also rectified by D1 to drive the meter circuitry.

The mixer gives stereo outputs — this is achieved by duplicating the circuitry for the second channel. The exception is the tone controls which are dual gang potentiometers. Note that the volume controls are individual units.

The output from each of the preamplifiers can be switched into a headphone mixer-amplifier to allow monitoring of any input. This enables the operator to cue inputs before they are faded. The only volume control for the headphones is on the rear panel as it is not normally adjusted. Again, if inputs of widely differing outputs are used the resistors between the volume control potentiometers and the switches can be changed.

The power supply is simply a full wave rectified supply with a centre tap giving about $\pm 12V$ dc.

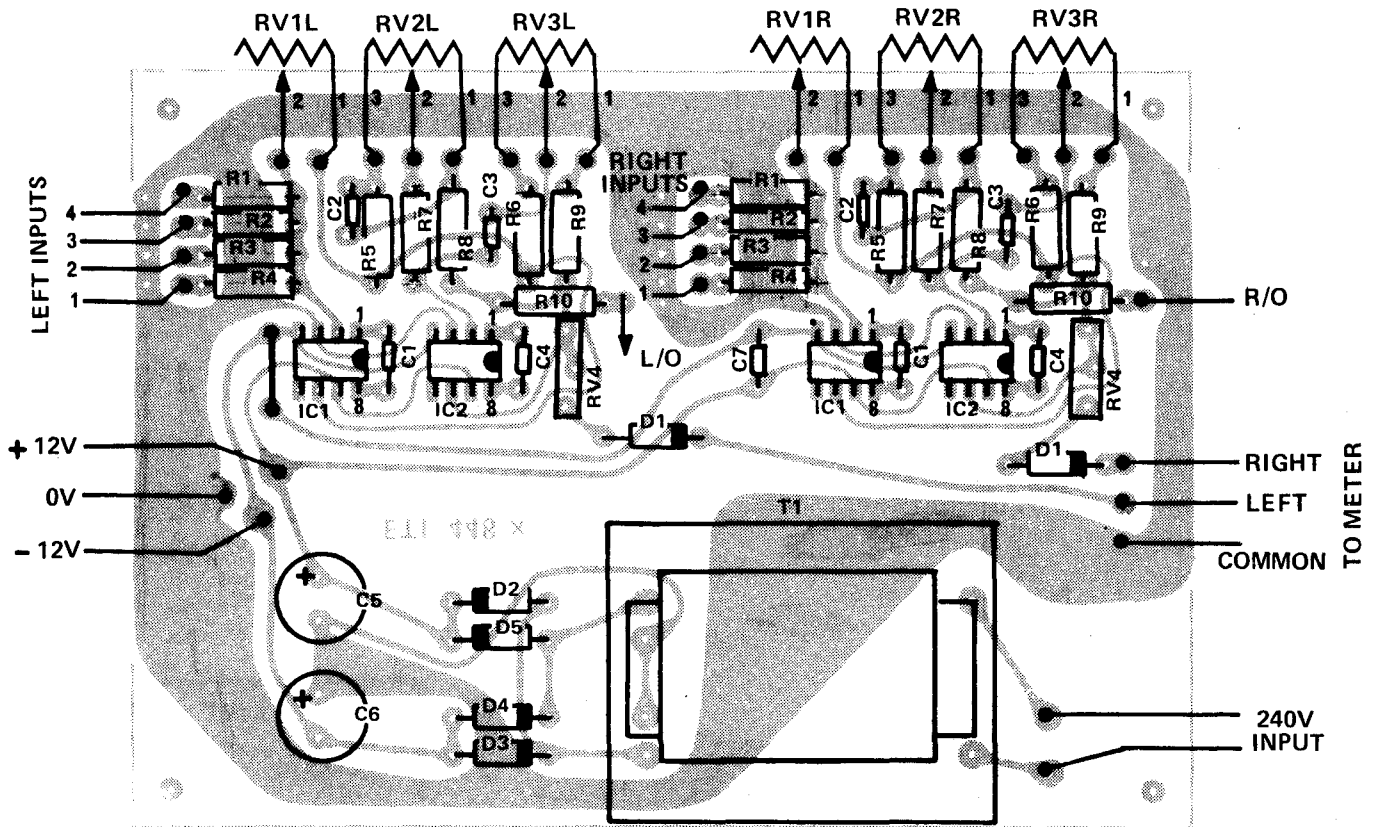


Fig. 3. Component overlay.

panel (as shown) or the bottom panel (if more room is needed). If operation in an electrically noisy environment is expected it may pay to line the box with "Alfoil" or similar and earth it.

The wiring of the boards can be done using the overlays or the wiring diagram.

When wiring the unit keep all the wiring as short as possible especially if

shielded cable is not being used. The 240 V wiring should be kept well away from the rest of the wiring to minimise pickup. For this reason we moved the power switch from the front panel (where we originally had it) to the rear panel, next to the power cable entry. The front and rear panels should also be earthed to minimise pickup of the 50 Hz signal.

With the headphone amplifier it is suggested that if only headphones are to be driven a 100 ohm resistor be fitted in the output. This will protect the headphones and also obviate the need for heatsinks on the LM380 (otherwise this output would give about 1 W into 8 ohms). This resistor can be fitted across spare terminals on the headphone sockets.

Fig. 4. Interconnection diagram of individual volume controls and headphone amplifier selector switches.

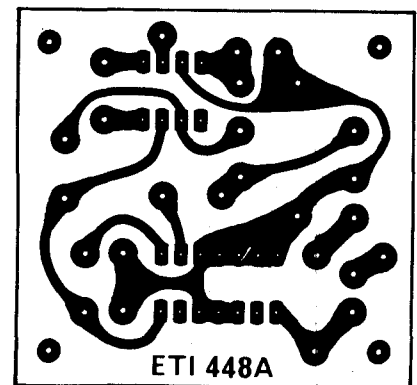
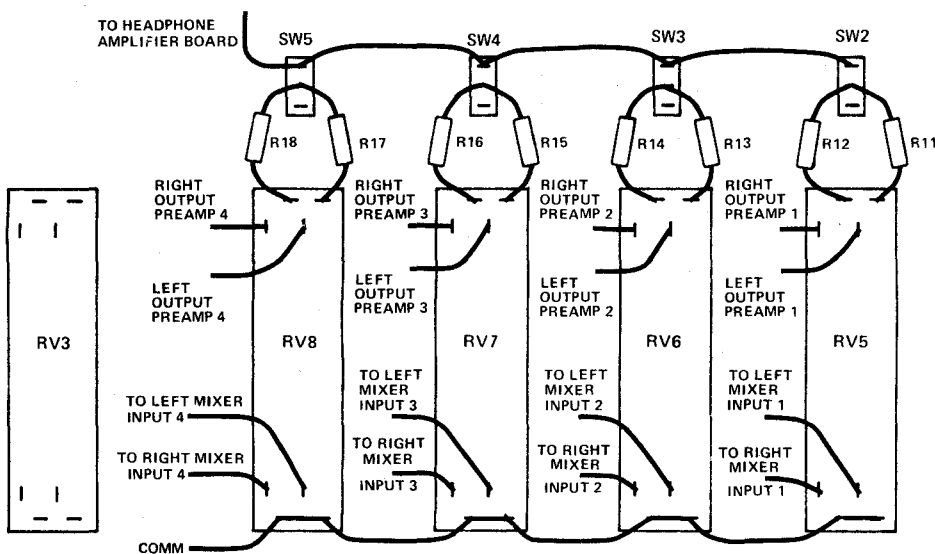


Fig. 5. Printed circuit layout of mixer. Full size 52 x 50 mm.

Project 448

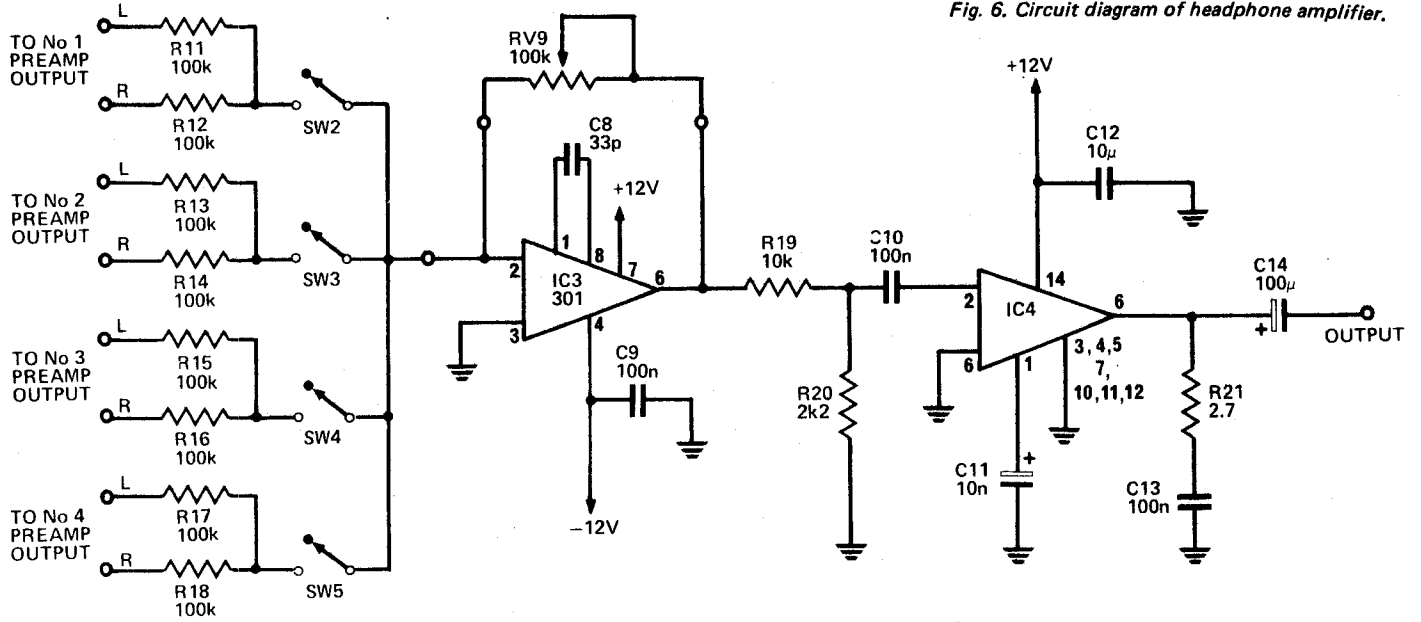


Fig. 6. Circuit diagram of headphone amplifier.

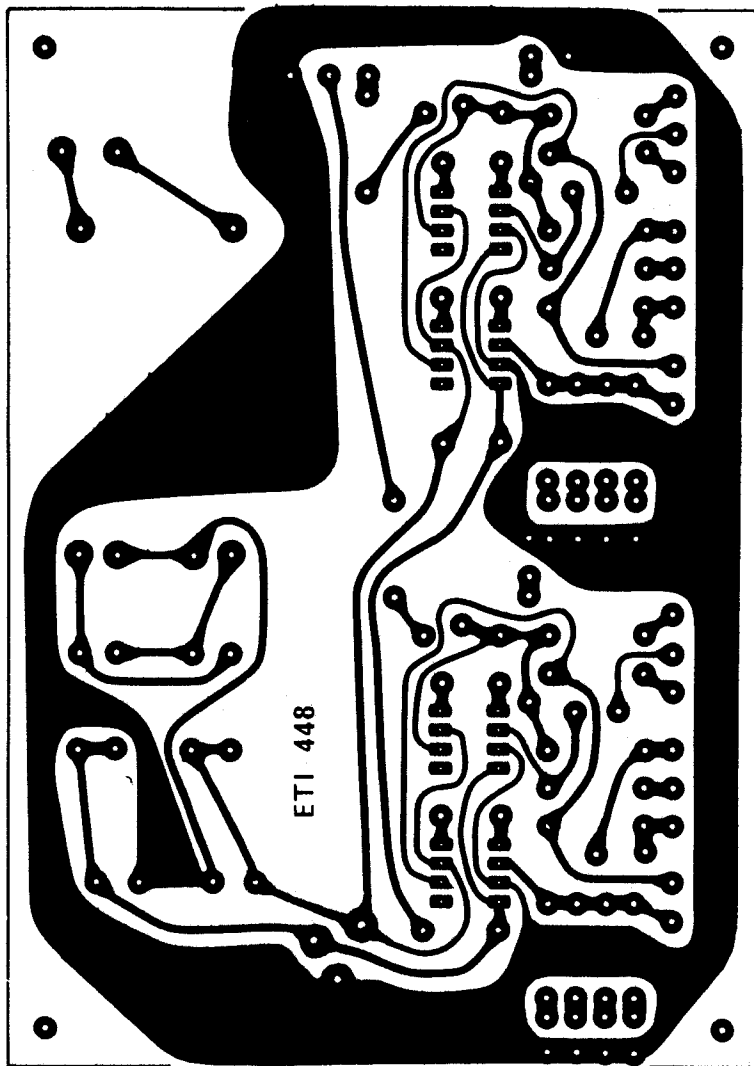


Fig. 7. Printed circuit layout of mixer and power supply board. Full size 140 x 100mm.

PARTS LIST - ETI 448A

Resistors all 1/2w 5%

R11-R18 100k
R19 10k
R20 2k2
R21 2.7 ohm

Potentiometer

RV9 100k log rotary

Capacitors

C8 33p ceramic
C9, 10 100n polyester
C11, 12 10µ 16 V electro
C13 100n polyester
C14 100µ 16 V electro

IC3 LM301A

IC4 LM380

SW2-SW5 single pole toggle

pc board ETI 448A

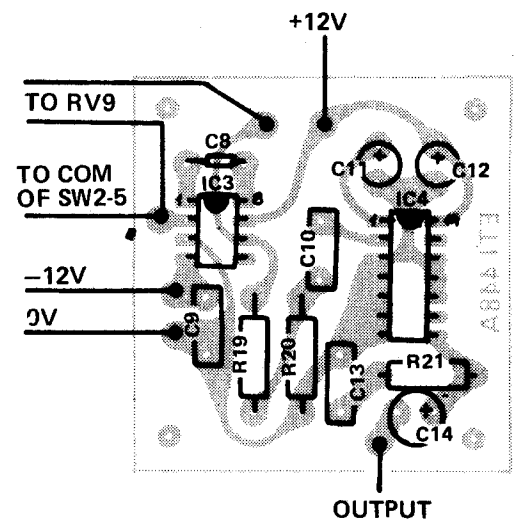


Fig. 8. Component overlay of headphone amplifier.

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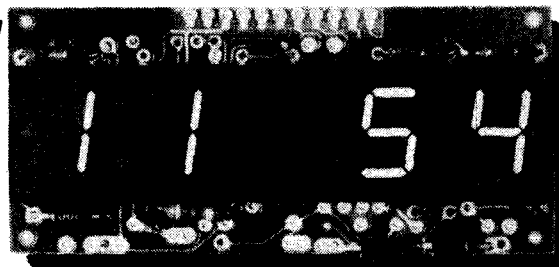
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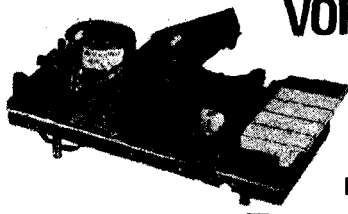
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SIMPLE INTERCOM

project electronics

This simple intercom uses the ETI 061 amplifier module, has no transformers, preset volume control and has the speakers doubling as microphones.

OUR SIMPLE AMPLIFIER PROJECT described in Electronics Today last month is a general purpose unit. It may be used just as described for innumerable experiments — or it may be used as the basis for very many other simple and practical projects, several of which will be included in this series.

Here is an intercom unit based on the simple amplifier plus a handful of further components. It is a versatile little unit that provides two-way communication over any distance ever likely to be required.

To save the cost of both microphones and loudspeakers we have used a single loudspeaker at each end and these double as microphones. Almost any 8 ohm or 16 ohm speakers may be used. Units of two to three inch diameter are readily available for not much more than a dollar or so. For the best performance these units should each be mounted in a small enclosure. The size and shape of the enclosure is not particularly important. An open backed box about the size of a shoe box would be fine.

The system may be converted for use as a baby crying alarm simply by wiring a toggle switch across the 'push-to-talk' push button at the desired station.

Construction

The ETI 064 intercom consists of the previously described 061 amplifier and a handful of extra components — the latter being mounted separately on printed circuit board, Veroboard or tag strip — as desired. The 061 amplifier should be assembled, as shown on page 62 of Electronics Today last month.

The two separate assemblies should then be interconnected as shown in Fig. 2. The two speaker switches should be located wherever convenient — either on the speaker enclosures or any position that is handy when using the units.

Three wires are used to connect the two speaker stations together. The wire used is not critical — bell wire or similar will be fine. Twist the wires together along their entire length to prevent picking up unwanted noise. About ten twists per metre is sufficient.

One nine volt supply is used to drive the complete unit — i.e. the simple amplifier draws into power from the battery used in the intercom. Note however that as the unit is 'on' all the time a nine volt battery eliminator should preferably be used in place of the battery shown. Suitable eliminators may be built following plans previously published in ETI or bought for a few dollars from almost any electrical supply store.

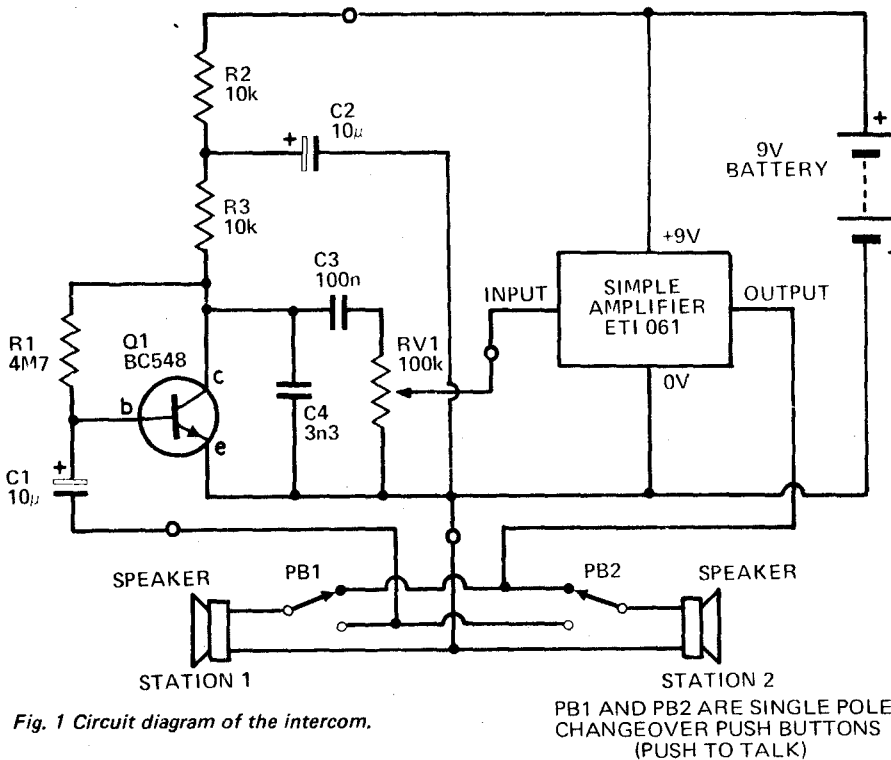


Fig. 1 Circuit diagram of the intercom.

How It Works ETI 064

Conventional moving coil loudspeakers double as speakers and microphones in each station. As they are not very efficient used like this we have added a single common emitter transistor amplifier to boost the output about 40 dB (100 times) before feeding it into the ETI 061 amplifier.

The collector load for Q1 is R2

and R3 with capacitor C2 to prevent supply voltage variations getting into the output. The dc biasing for Q1 is provided by R1 which causes the output of Q1 (collector) to stabilise at about 2V. The actual voltage depends on the β (gain) of Q1.

The output is adjustable by RV1 to provide adequate volume without incurring feedback.

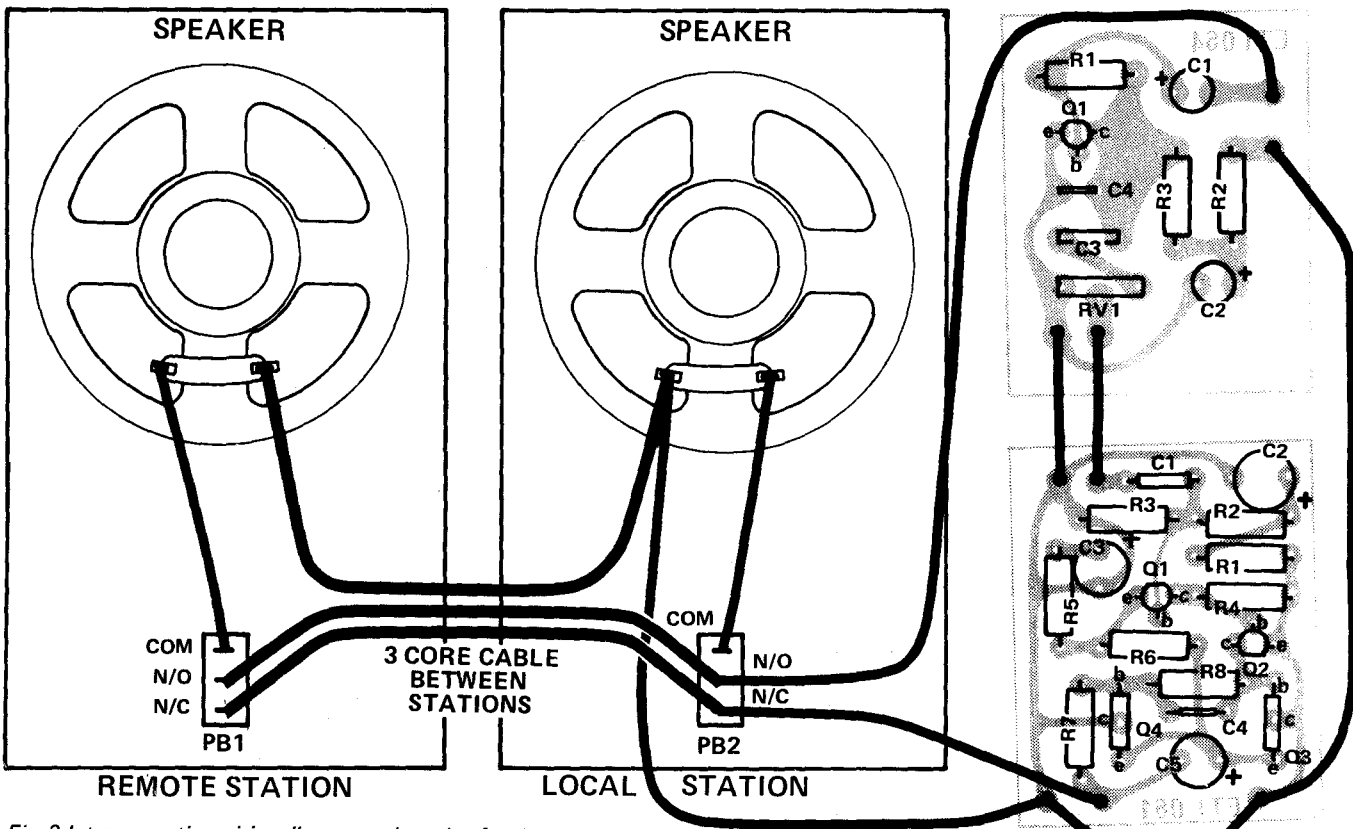


Fig. 2 Interconnecting wiring diagram and overlay for the PC version of the unit. The overlay for the amplifier is also shown.

PARTS LIST – ETI 064

R1	Resistor	4M7	½ W	5%
R2,3	Resistors	10 k	½ W	5%
RV1	Potentiometer	100 k	trim type	
C1,2	Capacitors	10 μF	16 V electro	
C3	Capacitor	100 nF	polyester	
C4	Capacitor	3n3	polyester	
Q1	Transistor	BC548		

2 push buttons with single changeover contact
 2 speakers 8 or 16 ohm
 PCB ETI 064 or Veroboard 1" x 2.5"
 Suitable boxes
 Amplifier module ETI 061

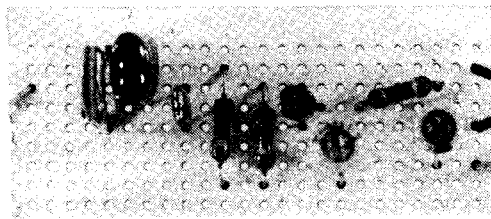


Fig. 4 Use this diagram to wire the Veroboard version of the unit.

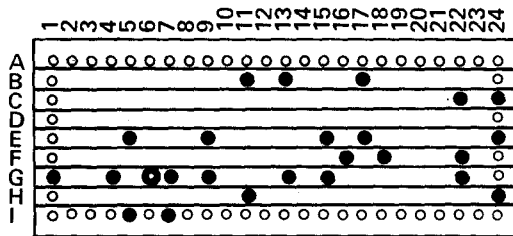
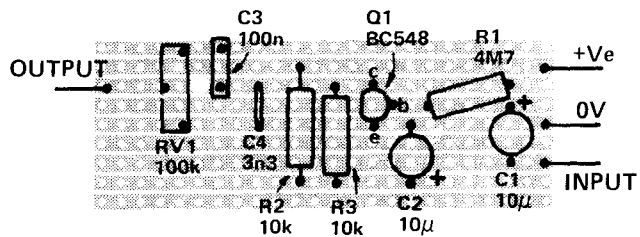


Fig. 5 Cut the veroboard track in the place indicated by the pad with the hole. The other dots are where soldered connections are made.

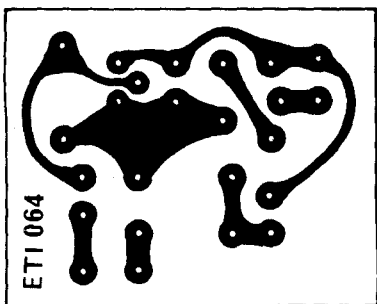
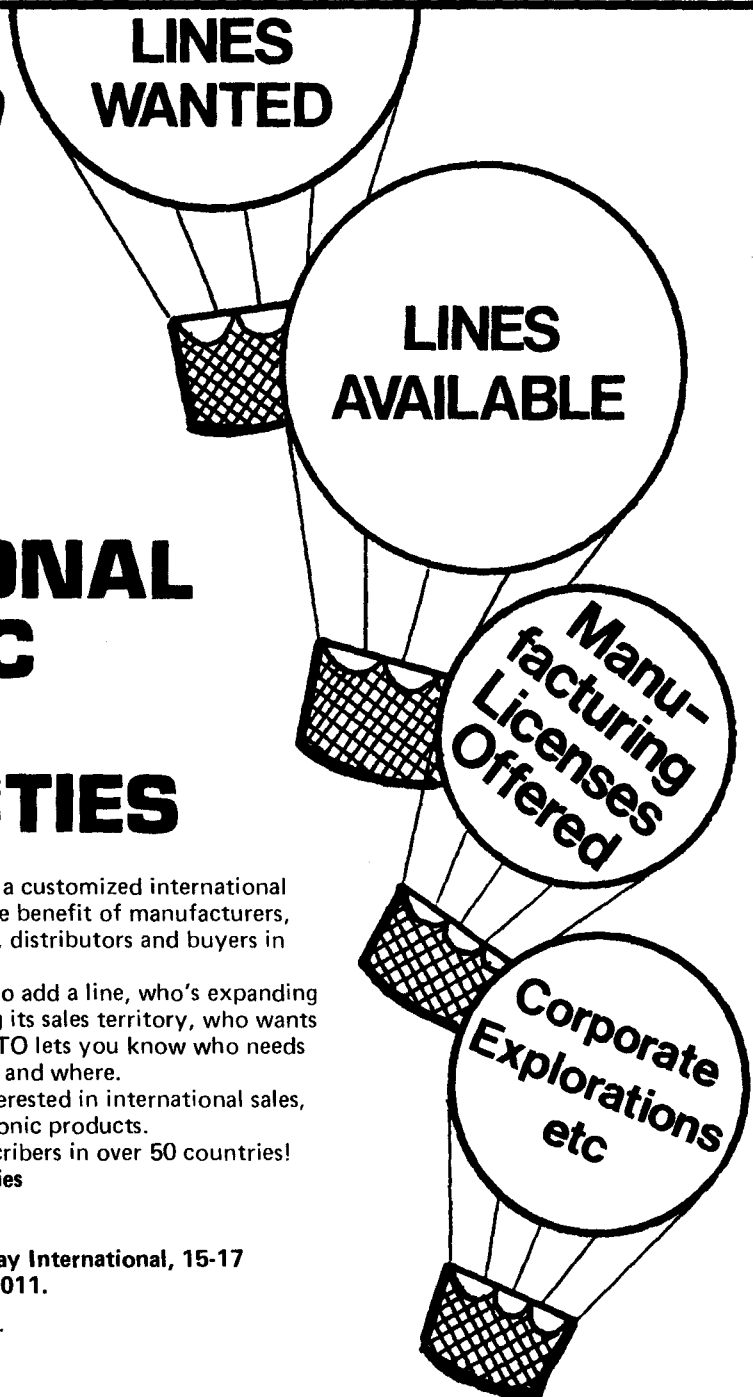


Fig. 3 Printed circuit board layout. Full size 50mm x 40mm.

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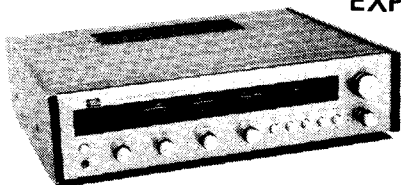
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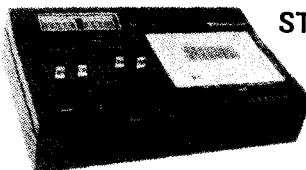


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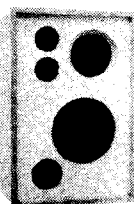
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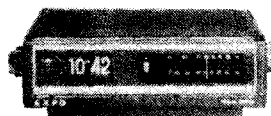
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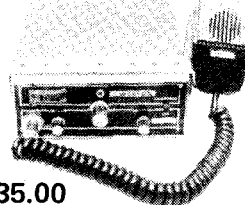


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AX10M	8 or 15	65-16,000	60	40	266	\$30.00
AX10M	8 or 15	40-11,000	30	60	311	\$67.50
402	8 or 15	30-3000	24	60	313x264	\$49.00
AUD10M	8 or 15	2KHz-22KHz	Dome Tweeter	40	112	\$25.00
200	8 or 15	1.4Hz-12KHz	Mid Range Horn	50	158x81x242	\$55.00
AXENT	8			60		\$45.00
100	8 or 15	20-2000		50		\$66.00
HI FAX	8 or 15	20-2000		50		\$66.00
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AUD10M	8 or 15					
12P	8 or 15					
AUD10M	8 or 15					
15P	8 or 15					
AUD10M	8 or 15					
18P	8 or 15					

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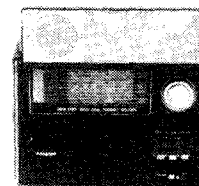
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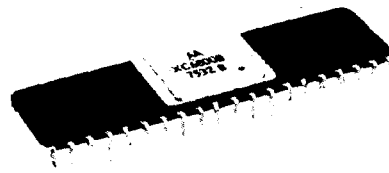
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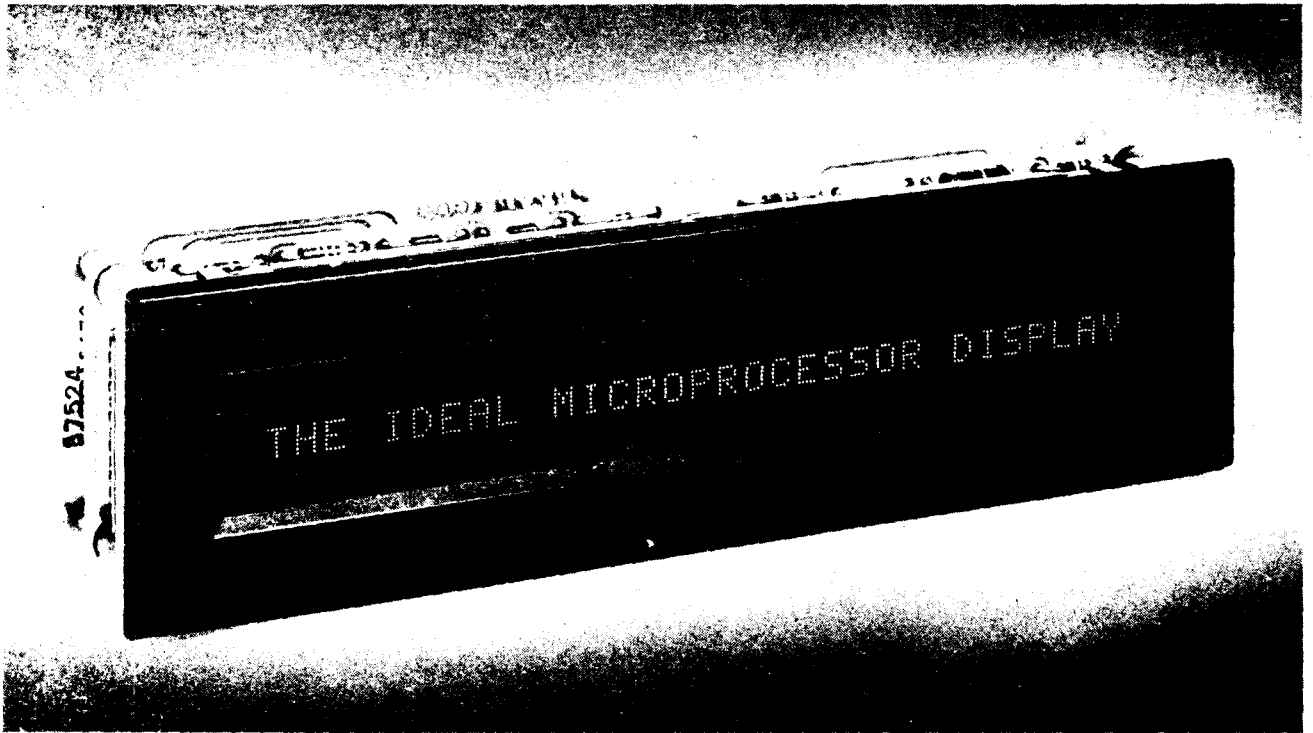


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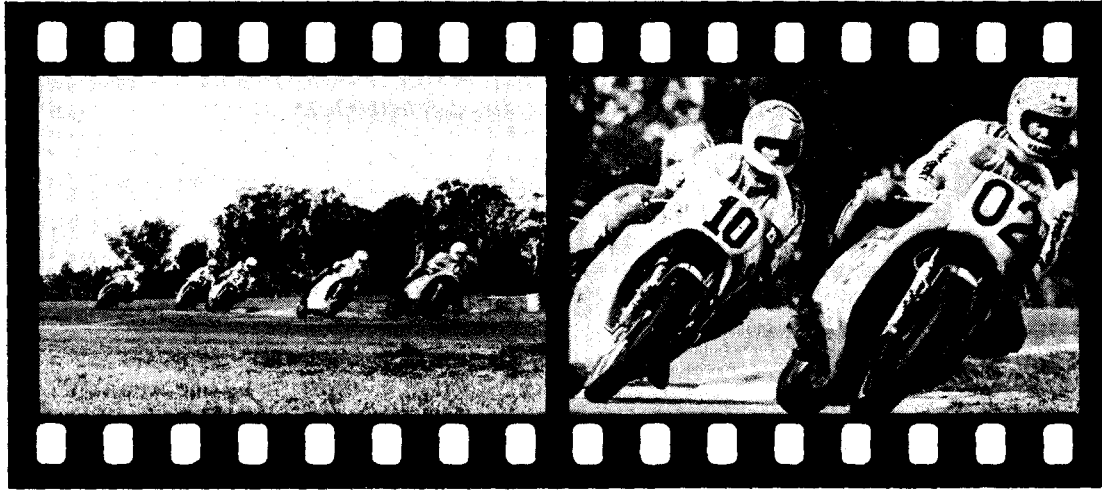


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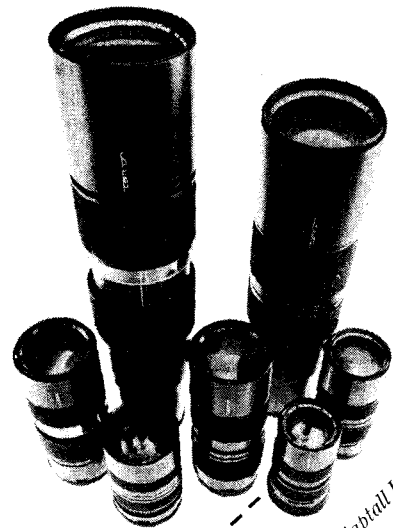
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Is that you?

-a computer that checks signatures

A growing need to check people's identity automatically has led to the development of a computer that verifies signatures by the speed and sequence of pen movements as well as by the finished sample — this report by R. S. Watson & P. J. Pobjee of Britain's National Physical Laboratory.

MODERN TECHNOLOGY HAS, ironically, increased the opportunities for crime and its rewards. Easier and more widespread facilities for getting goods on credit and the introduction of electronic fund transfer systems have made it possible to make money directly by fraud.

Nowadays, too, there are many places where people cannot be allowed to enter unless they are authorised. They may house stocks of valuable or dangerous material or stores of confidential information, often in the form of computer records. Providing guards to check people's identity costs a lot of money — there is a need for some

automatic system of checking that people are who they are supposed to be.

There are two ways of tackling the problem. First is the method of providing tokens such as credit cards or pass cards or even secret codes. But tokens can be lost or stolen or lent to other people. The second method is to make use of some human property such as fingerprints, body weight, or other physical dimension. Unfortunately, people often object to such things being used and, in any case, measurement can be expensive to automate. Together with voice 'prints' these visible attributes can still be imitated, another drawback.

PEN MOVEMENTS

Signing is the traditional method for authorising documents, and signatures represent a well practised human behaviour pattern. Although the visible mark can be easily copied or traced, the way in which it is written is also characteristic of the writer. This means that additional information can be obtained by measuring the *speed* and *sequence* with which the pen is moved across the paper.

It follows that, in any automatic system for recognising signatures as they are written, the first requirement is for an economic way of obtaining this hidden information without upsetting the writer's natural rhythm. This was obtained by inventing a simple electronic notepad that produced a sequence of electrical signals corresponding to the signing action without being connected to the writer's pen. This pad has been further developed commercially and is marketed by Quest Automation as a data entry device under the name Datapad.

The second stage was the study of a great number of signatures to choose a method of measurement that could ignore minor variations between samples from the same writer, while preserving his distinguishing features. Over 10 000 signatures were collected from more than 500 writers from all walks of life. When we examined these with a view to isolating the variables, four rather obvious factors emerged. These were name, style, context and noise.

The *name* forms the basic structure. It may be short, such as B. Nye or long, with 30 letters or more — Sir Frederick

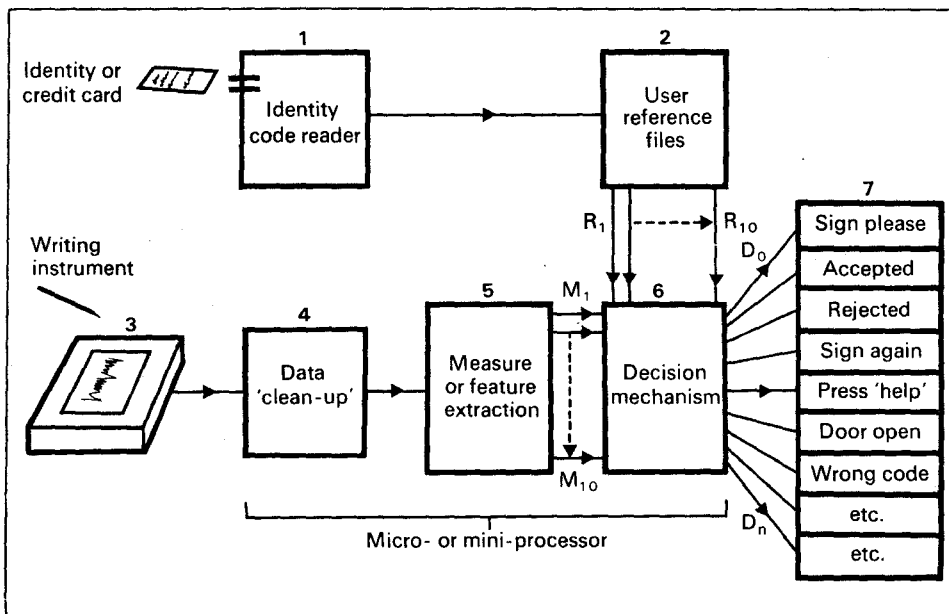


Fig. 1. Basic signature validation machine. The numbers are referred to in the text.

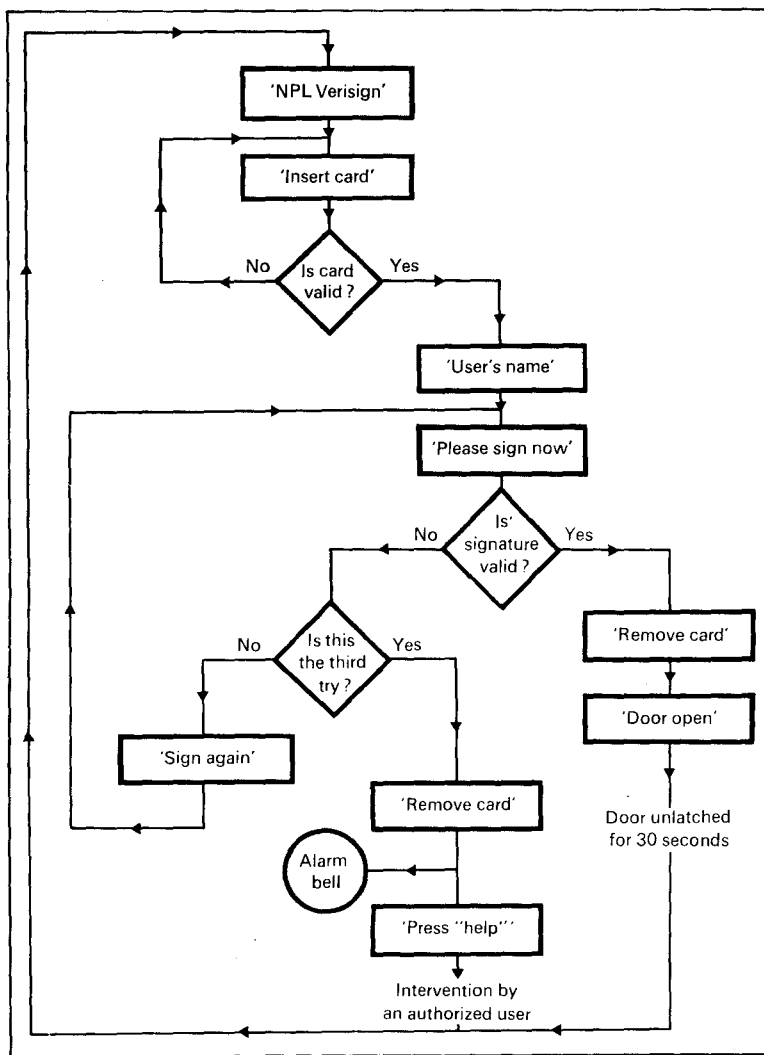


Fig. 2. Simplified flow chart of operation.

Marmaduke Bertwhistle. The name may be written in different languages, or scripts such as Roman, Russian, Arabic, Japanese, Hebrew or for that matter any well practised group of symbols. In some cases a person's initial are acceptable.

By *style* we mean the variations about the name form. Many people have a repertoire of styles which they use on various occasions. A number of common examples which we met were a 'working or everyday use' style, a 'cheque book' style and what might be called an 'impress the boss' style.

Context is the modification to a given style caused by what the individual is doing at the time. The rhythmic properties of a person's signature can vary according to his attitude to the transaction. The signing of an important document will affect the way he writes more than a trivial event such as the receipt of articles worth a few cents.

All the other influences that may

affect the signing behaviour we have called the *noise* factor. The weather may be included in this category and a number of signatures were collected from people arriving at the laboratory in midwinter. Other samples were obtained from people in various states of health. In one case drugs were being taken to alleviate the symptoms of a nervous condition. Then there is the 'after business lunch effect' which can influence the signing rhythms!

Our large data bank of signatures was supported by other experiences of interaction between man and machine. This enabled a team led by Dr J. Parks of NPL to develop powerful techniques to overcome many of the difficulties.

Peter Hawkes of the UK's National Research Development Corporation and Stephen Dennis of Inter-Bank Research Organisation had been following progress with interest, and a joint venture was formed between NRDC, INRO and NPL to construct a

prototype machine for VERIFICATION of SIGNatures (VERISIGN).

Diagram 1 illustrates the basic building blocks of the Verisign machine. A user first enters his personal identity code either through keyboard or badge reader (1). The code, which in our case is a four digit number, is used to extract the user's reference file (2) containing a set of 10 reference parameters (R1-R10). These are passed to the decision mechanism (6) and a request flashed to the output display (7) for the person to sign his name on the Datapad (3).

The Datapad has an electro-sensitive surface on which movements of the writing stylus are converted into a 'string' of interleaved x, y co-ordinates showing how far across and up or down particular points are. This 'data string' is then processed (4), to remove artifacts such as marks made accidentally by the user.

Analysis of the 'cleaned up' data occurs at (5) in which measurements are made on certain properties which characterise the signing pattern. Examples of possible measurements are the number of crossings made by the x or y co-ordinates over a datum line or the total time spent in writing. Many other functions of position and time may be chosen.

The properties or parameters can be selected 'locally', that is within certain areas, or 'globally', with the measurements taken over the whole signature.

Over 100 measures were tested for their ability to discriminate between writers, while remaining insensitive to each person's own variation. From these, ten measures were selected and used to generate the values M1-M10 which are passed to the decision mechanism (6). Here a comparison is made with those obtained from the claimed reference set (R1-R10). The degree of similarity or closeness of fit in relation to a set threshold value determines once of a number of decision (D1-Dn). A close fit, that is below the threshold value, is accepted. A poor fit causes the signature to be rejected and displays a request for further samples.

A hierarchy of decision procedures is used allowing 'context' factors such as customer importance or the value of the transaction to be incorporated. The

Is that you?

-a computer that checks signatures

decision mechanism can be easily organised in a number of different ways to suit individual requirements.

Establishing a set of measures to use as a reference for one person is a vital part of the smooth functioning of the machine. Security against impersonation, without the rejection of genuine attempts, will depend on how well the reference measures characterise the writer.

Anyone who will be using the machine is first asked to submit five specimen signatures. The spread of this group is then examined by the machine for any gross inconsistencies. Signatures that lie outside a given tolerance band are rejected and further samples requested to make up the number. The variation in the reference group (variability factor, VF) provides a useful means of assessing what the chances are for successful impersonation by unauthorised users. The lower this factor the higher the security and, of course, the reverse is true.

Knowledge of the degree of security is unknown to either the user or impersonator and in any case the rating value together with the reference list is updated each time a test signature is accepted. This updating mechanism can also keep track of long-term variation in the way a person writes his signature.

The basic flow chart of the Verisign machine is shown in diagram 2. Three attempts at writing a signature are permitted before some form of alert is given.

The computer program, apart from a few modules, is written in standard Fortran IV language and occupies about 12 000 words of core store. Twenty words are required for each person's reference parameters plus an extra 10 for performance logging.

We used a 16K mini-computer which provided reference file space for up to 120 people. The time to verify a signature was less than 100 milliseconds. This meant that a complete transaction, including the entry of a personal identity code, could be completed inside 20 seconds.

TESTS

The system was tested in various situations including remote operation over public telephone lines. In addition, two full-scale experiments were carried out. For the first, in the entrance hall at our laboratory, the participants identified themselves as they entered and left the building. The 71 people who took part included typists, security officers, members of the services, professional engineers and scientists. Out of 2000 attempts made at

identification by singing, 96 per cent were successful.

The second experiment controlled entry to the computer room of a different government establishment. Here, the 47 passholders, often carrying equipment or trays of cards, used the Verisign terminal over a period of several weeks. The results of this experiment were similar to the first.

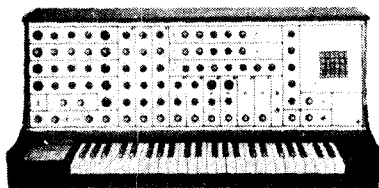
It is, of course, one thing to ensure that the genuine person is identified correctly with the minimum fuss or bother. It is another to prevent the less scrupulous their practising art! With this in mind, at the end of both experiments we displayed a number of target signatures and invited everyone to try his hand at copying them. With the first experiment at NPL, although one or two came very close, no-one was able to obtain a 'signature valid' signal. A lower threshold was used for the second experiment and the decision scores were displayed as an incentive. No limit was placed on the number of attempts, allowed and under these less rigorous, unrealistic conditions a few people were eventually successful.

No security system is perfect but the hierarchy of this one allows the degree of security to be balanced against the possibility of rejecting an authorised user. ●

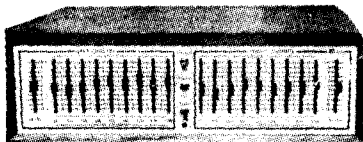
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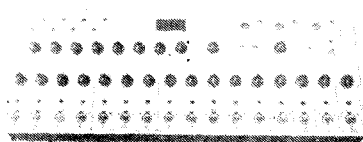
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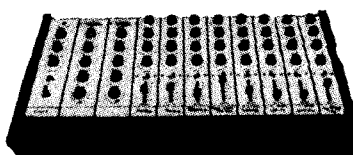
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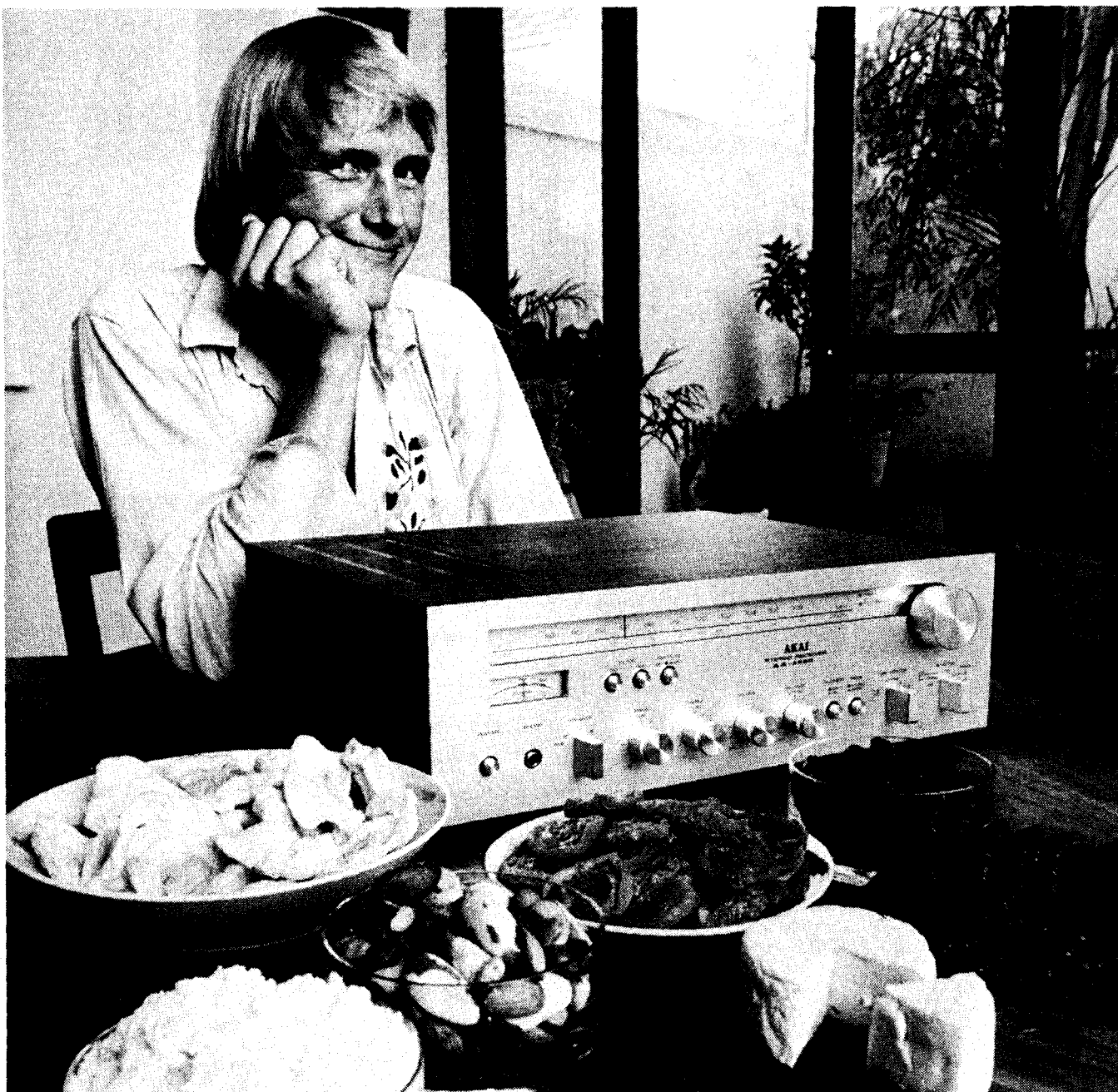
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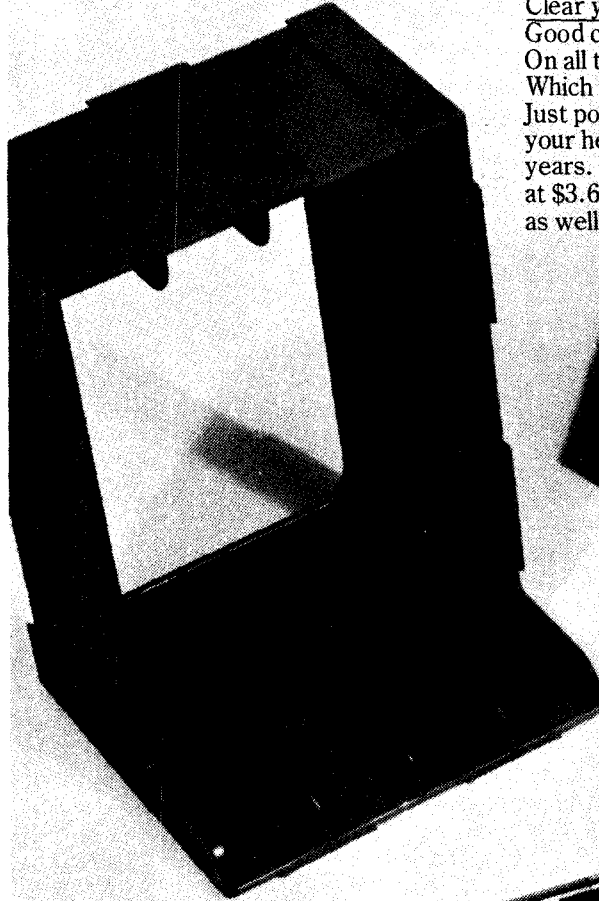
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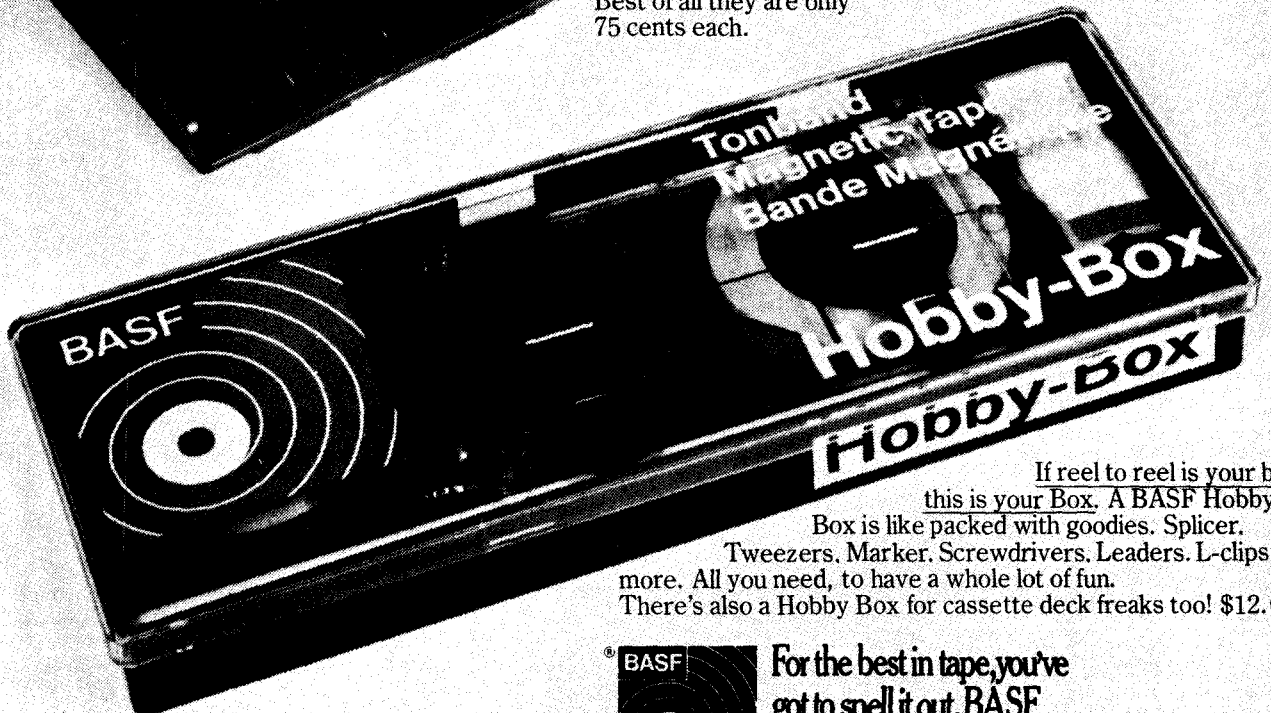
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CMOS

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DISPLAY DRIVERS					
WITH COUNTERS		FOR LCD DISPLAYS		FOR LED DISPLAYS	
MULTIPLEXERS DEMUTIPLEXERS ANALOG-DIGITAL DATA SELECTORS		PHASE LOCKED LOOPS	RAMs	QUAD BILATERAL SWITCHES	
ARITHMETIC CIRCUITS					
ADDERS & COMPARATORS		ALUs & RATE MULTIPLIERS	PARITY GENERATORS & CHECKERS	MULTIPOINT REGISTERS	

The headings we used to classify the CMOS devices. The lists starts on page 80.

TTL to CMOS. Functionally Equivalent Types

TTL	CMOS	TTL	CMOS	TTL	CMOS
7400	4011	7475	4042	74150	4067
7401	40107	7476	4027	74151	4051 4097
7402	4001	7477	4042	74152	4051 4097
7404	4009 4049	7478	4027	74153	4052
7406	4009 4049	7483	4008	74154	4514 4515
7407	4010 4050	7485	4063	74155	4555 4556
7408	4081	7486	4030 4070	74156	4555 4556
7410	4023	7490	4518	74157	4019
7411	4073	7491	4015 4094	74164	4015
7420	4012	7493	4520	74165	4021
7425	4002	7494	4035	74166	4014
7427	4025	7495	40104 40194	74167	4527
7428	4001	7499	40104 40194	74173	4076
7430	4068	74100	4034	74178	4035
7432	4071	74104	4095	74179	4035
7437	4011	74105	4095	74180	40101
7440	4012	74107	4027	74181	40181
7442	4028	74110	4095	74182	40182
7445	4028	74111	4027	74190	4510
7446	4511 4055	74121	4047 4098	74191	4516
7447	4511 4055	74122	4047 4098	74194	40104 40194
7448	4511 4055	74123	4098	74195	4035
7449	4511 4055	74125	4502	74198	4034
7450	4085	74126	4502	74200	4061
7453	4086	74132	4093	74251	4051 4097
7454	4086	74136	4030 4070	74279	4044
7470	4096	74141	4028	74283	4008
7472	4095	74145	4028	74290	4518
7473	4027	74148	4532	74293	4520
7474	4013				

CMOS

This is a list of commonly-available CMOS in the 4000 and 74C00 series (plus one or two other similar types). Refer to the plan on page 79 to see the headings we have used to classify the devices.

When you see a blob in front of a number this means that the device is available in LOCMOS. LOCMOS ICs are functionally identical to other CMOS but the outputs are buffered. This means that they cannot be operated in the linear mode (as amplifiers, etc). CMOS from Solid-State Scientific is also buffered.

SINGLE LEVEL GATES

NOR/NAND

- 4000 Dual 3-Input NOR Gate Plus Inverter
- 4001 Quad 2 Input NOR Gate
- 4002 Dual 4 Input NOR Gate
- 4011 Quad 2 Input NAND Gate
- 4012 Dual 4 Input NAND Gate
- 4023 Triple 3 Input NAND Gate
- 4025 Triple 3 Input NOR Gate
- 4068 8-Input NAND Gate
- 4078 8-Input NOR Gate
- 40107 Dual 2-Input NAND Buffer Driver
- 74C00 Quad 2-Input NAND Gate
- 74C02 Quad 2-Input NOR Gate
- 74C10 Triple 3-Input NAND Gate
- 74C20 Dual 4-Input NAND Gate
- 74C30 8-Input NAND Gate

OR/AND

- 4071 Quad 2 Input OR Gate
- 4072 Dual 4 Input OR Gate
- 4073 Triple 3-Input AND Gate
- 4075 Triple 3 Input OR Gate
- 4081 Quad 2 Input AND Gate
- 4082 Dual 4-Input AND Gate
- 74C08 Quad 2-Input AND Gate
- 74C32 Quad 2-Input OR Gate

BUFFERS & INVERTERS

- 4007 Dual Complementary Pair Plus Inverter
- 4009 Hex Inverting Buffer
- 4010 Hex Buffer
- 4041 Quad True Complement Buffer
- 4049 Hex Inverting Buffer
- 4050 Hex Buffer
- 4069 Hex Inverter
- 4502 Strobed Hex Inverter Buffer
- 4503 Hex Bus Driver
- 40097 3-S Hex Non-Inverting Buffer
- 40098 3-S Hex Inverting Buffer
- 40107 Dual 2-Input NAND Buffer Driver
- 74C04 Hex Inverter

- 74C901 Hex Inverting Buffer (TTL Interface)
- 74C902 Hex Buffer (TTL Interface)
- 74C903 Hex Inverting Buffer (MOS Interface)
- 74C904 Hex Buffer (MOS Interface)
- 74C906 Open Drain Buffer (Active Pull-down)
- 74C907 Open Drain Buffer (Active Pull-down)
- 74C908 Dual 30V Buffer
- 80C95 TRI-STATE Hex Buffer
- 80C96 TRI-STATE Hex Inverting Buffer
- 80C97 TRI-STATE Hex Buffer
- 80C98 TRI-STATE Hex Inverting Buffer
- 88C29 Quad Line Driver (Single Ended)
- 88C30 Dual Line Driver (Twisted Pair)

LEVEL SHIFTER

- 40109 Quad Low-to-High Voltage Level Shifter

MULTI-LEVEL GATES

MULTIFUNCTION & AOI

- 4019 Quad AND/OR Select Gate
- 4030 Quad EX OR Gate
- 4037 Triple AND/OR Bi-Phase Pairs
- 4048 Expandable 8-Input Gate
- 4070 Quad Exclusive OR Gate
- 4077 Quad Exclusive NOR Gate
- 4085 Dual 2 Wide 2-Input AND/OR INVERT Gate
- 4086 Expandable 4-wide 2-Input AOI
- 4507 Quad EX OR Gate (74C86)
- 74C86 Quad EX-OR Gate

DECODERS & ENCODERS

- 4028 BCD to Decimal Decoder
- 4514 4-Bit Latch/4 to 16 Line Decoder (High)

- 4515 4-Bit Latch/4 to 16 Line Decoder (Low)
- 4532 8-Input Priority Encoder
- 4555 Dual 1-of-4 Decoder/Demultiplexer (Outputs High)
- 4556 Dual 1-of-4 Decoder Demultiplexer (Outputs Low)
- 74C42 BCD-to-Decimal Decoder
- 74C48 BCD-to-7 Segment Decoder
- 74C154 4 to 16 Line Decoder
- 74C915 7-Segment to BCD Decoder
- 74C922 16 Key Keyboard Encoder
- 74C923 20 Key Keyboard Encoder

SCHMITT TRIGGERS

- 4093 Quad 2-Input NAND Schmitt Trigger
- 40106 Hex Schmitt Trigger
- 4584 Hex Schmitt Trigger (74C14N)
- 74C14 Hex Schmitt Trigger
- 74C914 Hex Schmitt Trigger

MULTIVIBRATORS

FLIP-FLOPS & LATCHES

- 4013 Dual D Flip-Flop
- 4027 Dual JK Flip-Flop
- 4042 Quad D Latch
- 4043 Quad 3 State NOR R/S Latch

- 4044 Quad 3 State NAND R/S Latch
- 4076 TRI-STATE Quad D Flip-Flop
- 4095 Gated J-K Non-Inverting Inputs
- 4096 Gated J-K Inverting and Non-Inverting Inputs
- 4099 8-Bit Addressable Latch

- 40174 Hex D Flip-Flop
- 40175 Quad D Flip-Flop
- 4508 Dual 4-Bit Latch
- 4724 8-Bit Address Latch
- 74C73 Dual JK Flip-Flop
- 74C74 Dual D Flip-Flop
- 74C76 Dual JK Flip-Flop
- 74C107 Dual JK Flip-Flop
- 74C173 TRI-STATE Quad D Flip-Flop
- 74C174 Hex D Flip-Flop
- 74C175 Quad D Flip-Flop

ASTABLES & MONOSTABLES

- 4047 Monostable-Astable Multi-vibrator
- 4098 Dual Monostable Multi-vibrator
- 4528 Dual Retriggerable/Resettable Monostable Multivibrator
- 74C221 Dual Monostable Multi-vibrator

REGISTERS

SHIFT, STORAGE & FIFO

- | | |
|--|---|
| <ul style="list-style-type: none"> ● 4006 18-Bit S/R ● 4014 8-Bit S/R ● 4015 Dual 4-Bit S/R ● 4021 8-Bit S/R ● 4031 64-Bit S/R 4034 8-Bit S/R ● 4035 4-Bit P-In P-Out S/R 4094 8-stage Shift-and-Store Bus Register 40100 32-Bit Left/Right S/R 40104 3-state 4-Bit Left/Right S/R | <ul style="list-style-type: none"> ● 40194 4-Bit Left/Right S/R ● 40195 b-Bit Universal S/R ● 4076 TRI-STATE Quad D Flip-Flop 4099 8-Bit Addressable Latch 40108 4x4 multiport register 40105 4-word 4-Bit FIFO Buffer 74C164 8-Bit S-In P-Out S/R 74C165 8-Bit P-In S-Out S/R 74C195 4-Bit Parallel S/R 74C95 4-Bit R-S L-S Register |
|--|---|

COUNTERS

- | | |
|---|---|
| <ul style="list-style-type: none"> ● 4017 Decade Counter Divider ● 40108 Presettable Divide by "N" Counter ● 4020 14-Bit Ripple Carry Binary Counter/Divider ● 4022 Divide by 8 Counter Divider ● 4024 7-Bit Binary Counter ● 4029 Presettable Up/Down Counter 4040 12-Bit Binary/Ripple Counter 4045 21-stage Clock Timer 4059 Programmable Divide-by N Counter 4060 14-Stage Ripple Carry Binary ● 4510 BCD Up/Down Counter ● 4516 Binary Up/Down Counter ● 4518 Dual BCD Up Counter ● 4520 Dual Binary Up Counter 40102 BCD Presettable 8-Bit down Counter 40103 Binary presettable 8-Bit down Counter | <ul style="list-style-type: none"> ● 40192 Sync Up/Down Decade Counter ● 40193 Sync Up/Down Binary Counter ● 40160 Sync Decade Counter ● 40161 Sync Binary Counter ● 40162 Fully Sync Decade Counter ● 40163 Fully Sync Binary Counter 74C90 Decade Counter 74C93 4-Bit Binary Counter 74C160 Sync Decade Counter 74C161 Sync Binary Counter 74C162 Fully Sync Decade Counter 74C163 Fully Sync Binary Counter 74C192 Sync Up/Down Decade Counter 74C193 Sync Up/Down Binary Counter 74C925 4 Decade Counter Divider 74C926 4 Decade Counter Divider with Display Select 74C927 Minutes and Seconds Counter Driver 74C928 2-1/2 Decade Counter Driver |
|---|---|

DISPLAY DRIVERS

WITH COUNTERS

- 4026 Decade Counter/Divider with Seven-Segment Display outputs and Display Enable.
- 4033 Decade Counter/Divider with 7-Segment Display Outputs and Ripple Blanking.

FOR LCD DISPLAYS

- 4054 4-Line Liquid Crystal display Driver.
- 4055 BCD-to-Seven-Segment Decoder/Driver with 'Display-Frequency' Output, for Liquid Crystal Displays.



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DISPLAY DRIVERS cont'

- | | | |
|-------|---|-------------------------|
| 4056 | BCD-to-Seven-Segment Decoder/Driver with Strobed-Latch function, for Liquid Crystal Displays. | FOR LED DISPLAYS |
| ●4511 | BCD-to-7-Segment Latch/Decoder/Driver | |

MULTIPLEXERS/ DEMULTIPLEXERS ANALOG-DIGITAL DATA SELECTORS

- | | | |
|-------|--|--------------------------------------|
| ●4016 | Quad Bilateral Switch | |
| ●4019 | Quad AND/OR Select Gate | 40257 |
| ●4051 | Single 8 Channel Multiplexer | plexer (Outputs Low) |
| ●4052 | Differential 4 Channel Multiplexer | 74C150 |
| ●4053 | Triple 2 Channel Multiplexer | 74C151 |
| ●4066 | Counter Divider and Oscillator | 74C157 |
| ●4067 | 16-channel Multiplexer/ Demultiplexer | 82C19 |
| 4097 | Differential 8-channel Multiplexer/Demultiplexer | 74C150 16 to 1 Multiplexer |
| ●4519 | 4-Bit AND/OR Selector | 74C151 8-Channel Digital Multiplexer |
| ●4539 | DUAL 4-Input Multiplexer | 74C157 Quad 2-Input Multiplexer |
| ●4555 | Dual 1-of-4 Decoder/Demultiplexer (Outputs High) | 82C19 TRI-STATE 16 to 1 Multiplexer |
| ●4556 | Dual 1-of 4 Decoder/Demulti- | |

PHASE LOCKED LOOP

- 4046 Phase Locked Loop

RAMs

- | | | | |
|--------|--|--------|------------------------------|
| 4039 | 4-word, 8-bit, direct word-line addressing | 74C920 | 1 k (256 x 4) RAM |
| 4061 | 256-word by one bit static RAM | 74C921 | 1 k (256 x 4) Common I/O RAM |
| 4036 | 4-word, 8-Bit, binary-addressing | 74C929 | 1 k 16-Pin RAM |
| 40108 | 4x4 multiport register | 74C930 | 1 k 18-Pin RAM |
| ●4720 | 256 x 1 BIT RAM (15V) | | |
| ●4721 | 256 x 4 BIT RAM (15V) | | |
| 74C89 | 64-Bit RAM | | |
| 74C200 | 256-Bit 256 x 1 (RAM) | | |
| 74C910 | 256-Bit (74 x 4) RAM | | |

QUAD BILATERAL SWITCHES

- 4016 Quad Bilateral Switch
- 4066 Counter Divider and Oscillator

ARITHMETIC CIRCUITS

ALUs & RATE MULTIPLIERS

- 4057 4-Bit ALU
- 4089 Binary Rate Multiplier
- 4527 BCD Rate/Multiplier
- 40181 4-Bit ALU
- 40182 Look-ahead carry block

- 4030 Quad EX OR Gate
- 4070 Quad Exclusive OR Gate
- 4077 Quad Exclusive NOR Gate
- 74C83 4-Bit Binary Full Adder
- 74C85 4-Bit Comparator
- 74C909 Linear Comparator

ADDERS & COMPARATORS

- 4008 4-Bit Full Adder with Parallel Carry
- 4032 Triple Serial Adder (+logic)
- 4038 Triple Serial Adder (-logic)
- 4063 4-Bit Magnitude Comparator

PARITY GENERATORS & CHECKERS

- 40101 9-Bit parity generator and checker

MULTIPOINT REGISTERS

- 40108 4x4 multiport register

GET THE PICTURE?

ROLLEI 140RES PROFESSIONAL.

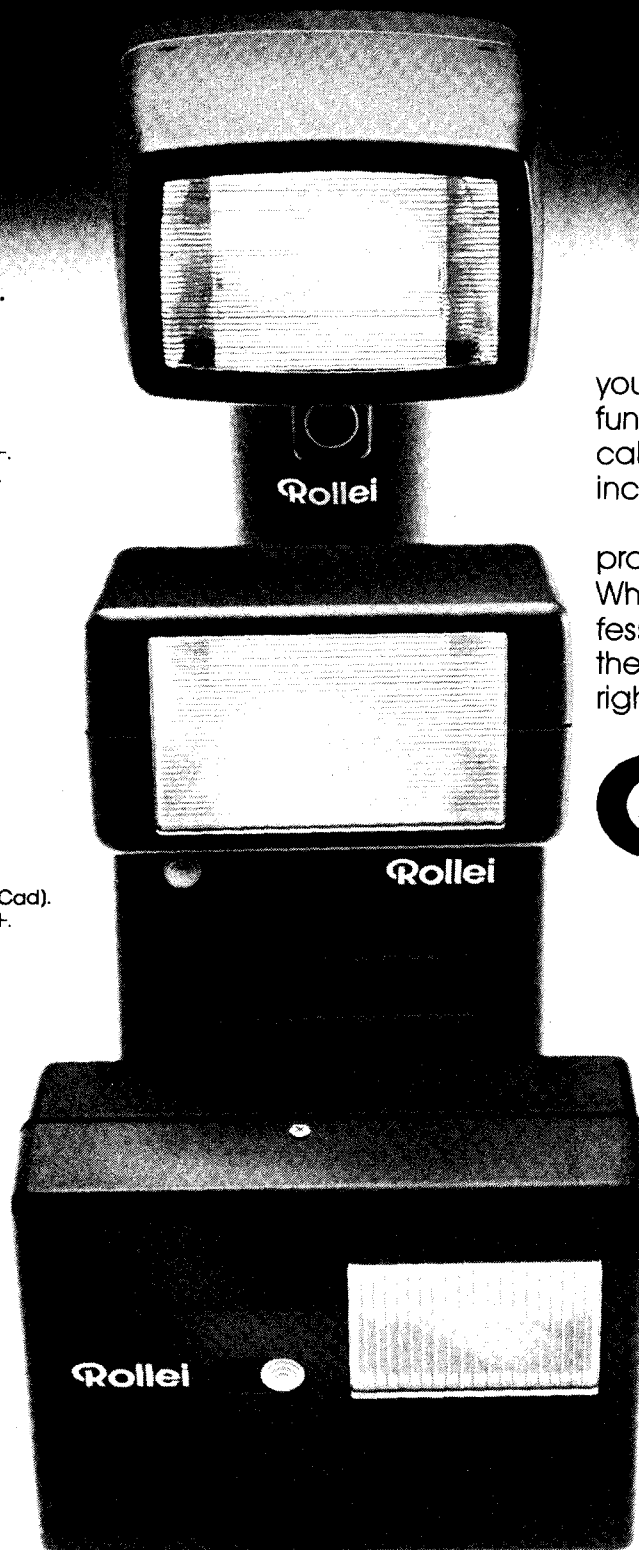
Powerful professional unit with swivel head for bounce flash.
GN: 130 (100ASA).
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DURATION: 1/400 to 1/20,000.

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RECHARGING TIME: 4 hours (Ni Cad).
FLASHES PER CHARGE: 70-1500+.
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DURATION: 1/400 to 1/20,000.

ROLLEI 128BC.

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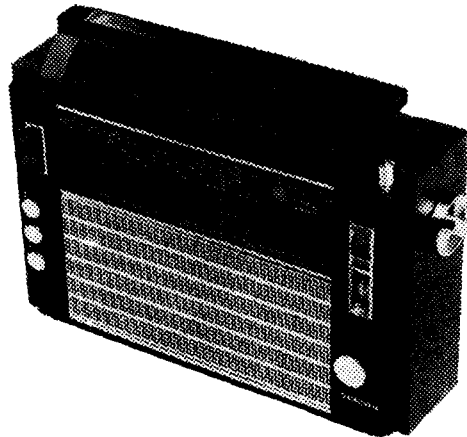
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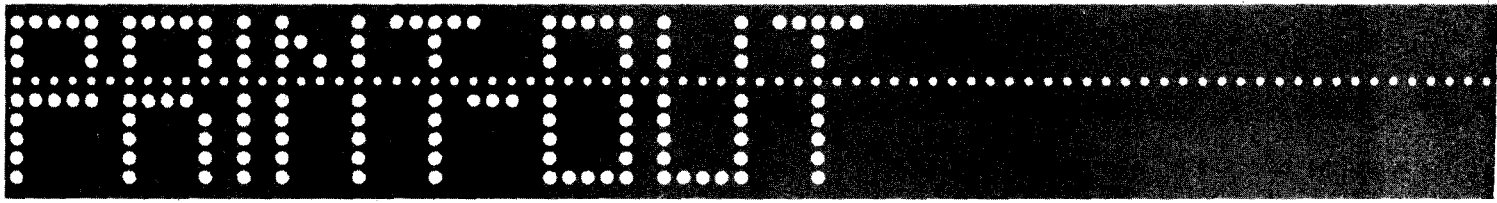
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Brisbane branch opening shortly.

MAIL ORDERS HANDLED PROMPTLY



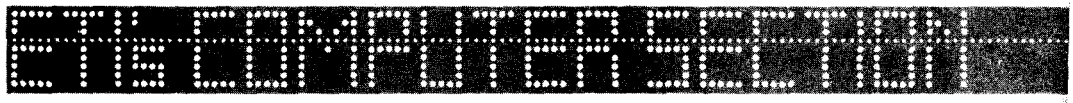
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THIS IS A NEW MONTHLY SECTION FOR ALL COMPUTER ENTHUSIASTS



This is the first edition of our new computer section so we want lots of feedback telling us what you want in future issues. Please write in if you have any problem, any suggestions, any technical hints . . . or any criticisms!

We're interested in hearing from companies — we want to know about any new developments or products. And if any of you have any interesting projects or software you want published we'd be keen to hear from you too.

We prefer to get typed letters (but we'll accept those written with a thumbnail dipped in tar!).

STARTING THIS MONTH ETI HAS a new section to cater for the increasing number of people involved with micro-computers. By now most readers are aware of the revolution in the electronics industry caused by the availability of microprocessors. Not since the development of the transistor in 1948 has a new product offered such exciting promise (or possibly frustration — if people try to ignore it, as some did with the transistor).

The computers built around micro-processors are already pushing into areas where previous computer technology has been unable to go. One such area is the hobby market. Costing between 100 and 1000 dollars, these machines are finding their way into more and more homes where people have discovered the delights and benefits of computing. While some machines are used just for fun, many are used to enhance another hobby: a more realistic model train system or an automatic ham station.

Another reason to become involved in home computing is to keep abreast of changes occurring in the electronics industry.

ETI's CONTRIBUTION

Watching the rapid emergence of this new technology it became clear that ETI would have to become involved, eventually, in computing. The question was how we could best contribute to this new field. Obviously a successful magazine gives its readers what they want, so the question became what do the readers want.

After considerable thought it was decided to provide a self-contained section within ETI with the aim of assisting you, the reader, to get the most enjoyment possible out of this new and exciting pastime, or to get the maximum benefit for your company if you are employed in the electronics industry. As time passes, computing will most certainly mature and change, and we intend that *PRINT-OUT* should develop along with it. But we intend to cover all aspects of computing and to give a balanced presentation of any interesting material.

USING MICROCOMPUTERS

Well then just what do we mean by interesting material? To give you an idea let's look at some general topic headings, but first let me make this point. The really exciting thing about computing is that the hardware is cheap enough and simple enough for you to buy and use yourself. This means your involvement can go much deeper than just reading about it and *wishing* you could do it to. And I think because of the computer's great versatility the excitement means, doing just about anything.

Consider the examples of an electric organ and a train controller. Once built each has only one use: the organ produces music and the controller controls trains. The computer can do both (a clever programmer can even make it do both at the same time). Some extra hardware is required for the interfacing but it is small compared to the amount used in a dis-

crete hardware design. The computer offsets this disadvantage with its ease of program change.

A difficult change in the hardware-designed controller is a relatively simple change to the program in the computer.

HARDWARE

There will be many readers who want to build their systems from scratch, or at least they'll want to do their own modifications. These are the people who enjoy building things themselves and they realise the advantages of less cost and more experience. Quite often a system purchased already assembled will need some hardware modification to allow for expansion. So we intend to publish articles on the use and misuse of hardware components. Of course you will also be kept up-to-date on the latest hardware releases and how to use them.

SOFTWARE

After having spent a lot of money and time getting your system into working condition, what happens next? You start feeding it software.

Well, yes, if you have a program. But before you have a program you have to write one, or get one of them from somewhere else. Unfortunately although most of us are born with the potential to be able to write programs we all have to learn how first, and this takes time.

It is generally accepted that writing good programs is an art; this means to

PRINT-OUT

get the best out of your system, you need to go beyond just learning the computer's instruction set. To make this happen for you we intend to include in *PRINT-OUT* features on programming as well as practical software hints and ideas.

EQUIPMENT REVIEWS

Many people prefer to buy their computer systems already assembled, or at least in kit form. By doing this they can save valuable time and reduce the possibility of something going wrong during the construction. Those outside the electronics industry who have little knowledge of electronics might also like to get into computing. Overseas many such enthusiasts are catered for by companies that specialise in hobby systems that need only 10 minutes reading and plugging into a wall socket to get going . . . Thus we see a place in *PRINT-OUT* for reviewing computer products in both assembled and un-assembled form. Such reviews take on added emphasis when one considers that for a while to come those buying these systems will be pretty unfamiliar

with computers and could make an unsuitable purchase. Already offered on the Australian market are several designs from overseas and we hear on the grapevine that there are to be released a couple of local designs. We are hoping to be able to bring you some interesting and revealing reviews in coming months.

I/O

Once the computer is going and the program is written, the most important part of the system becomes the interface. Sometimes its limitations will force you to go back and modify the program (sometimes even the hardware). Like a weak link in a chain no matter how fast or how powerful the computer is, without an effective interface the system is severely degraded. Computers are at their best when they do something, and if what you want to do is not explained in the supplied application note, quite likely the interface will become your own little problem. The interface is an important subject that will receive special attention in *PRINT-OUT*.

CONTRIBUTIONS

The greatest potential (and probably the most interesting aspect) of *PRINT-OUT* will be the exchange of ideas. Ideas on all areas of computing. We hope to make these exchanges a two-way happening, not only from us to you, but from you to us, so we can pass your ideas on to other computer enthusiasts. Often the problem you have just solved has been bugging another enthusiast for months. He might have even reached the stage where he is ready to give up or smash his computer. And all it could take is a short note from you to straighten him out. Chances are he has also solved a problem that's had you stumped. So with a little bit of co-operation you both benefit.

This co-operation needs one more thing, common ground. Somewhere where you can find his solution and he yours. We want *PRINT-OUT* to be such common ground, so if you have some neat solution to a tricky problem, write us a short letter describing what you have achieved. If it looks good to us we will show it to the world!

PRINT-OUT

This is our new section for users of microcomputer systems

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15 Boundary St, Rushcutters Bay
NSW 2011.**

Don't be left behind!

ETI COMPUTER SECTION

THE STORY OF THE HOBBY COMPUTER

OVER THE LAST FEW MONTHS WE have used the word "computer" quite a lot. PRINT-OUT is very much concerned with small computers and how their owners use them. Unfortunately, to many readers the word computer conjures up multiple images of machines ranging from slide rules to IBM 370s.

Because they are sufficiently powerful and versatile, many professional computer users prefer to buy a ready-made microcomputer rather than have the expense and problems of designing and manufacturing their own. These machines are available to the amateur enthusiast, but he is normally limited to buying a machine made specifically for the hobby market. This article looks at how these machines have developed and how companies have made available kits and low-cost systems to meet the demand.

How the hobby started

Historically it must be pretty fair to say that there have been individuals playing with computers since they first became a practical reality some 25 years ago. These would have been the lucky few who had access to the commercial computers: the engineers, operators and programmers employed to keep the machines running.

Although membership to this group must have been small, there were enough to support a newsletter — 1976 marks the 10th anniversary of the 'Amateur Computer Society Newsletter', which claims to be the oldest computer hobby publication in the world. But the big explosion came in early 1974 with the microprocessor, so that by late 1974 there were three microcomputer kits on the market, the Scelbi 8H, Radio-Electronics Mark 8 and the RGS 008A.

These were different to earlier designs (that had drifted into obscurity) because they offered reasonably powerful performance at a low price. Then at the beginning of 1975, thanks to the availability of the Intel 8080 micro-

processor chip, came the Altair 8800. Because of the formidable performance of the 8080, the Altair computer established itself as the T-model Ford of the hobby computer industry. That was 18 months ago and since then there have been a number of refinements to the hobby computer.

The early machines

The early microprocessor-based computers were very much scaled-down versions of the mini-computers that appeared in the mid-60s. Programs and data were entered one bit at a time from toggle switches mounted on the front panel. This meant a program of 20 bytes required something like 100 key toggles (+9 to set up the initial address). This unfortunately got more tedious as the program became longer (5000 for 1000 byte program!).

Many users went to first loading a short program (called a bootstrap), of no more than 20 bytes, that would carry out the loading of larger programs (making the computer do the work). The main program would then be kept on magnetic or paper tape. This saved much wear and tear on the fingers, but it meant you had to first obtain a cassette interface or paper tape reader and punch and you still had to load the main program at least once so that you could get it onto tape.

One other problem with this switch register method was the difficulty of outputting, for inspection, the contents of the microprocessor's internal registers. On the early mini-computers this was no problem because the CPU was made up of many chips and it was easy to run a few wires with lamps on the end to the particular registers. Unfortunately the microprocessor has only a limited number of pins and it is impossible for each register to have its own output for display purposes.

The only way around this problem is to load a special little program into memory that will output the contents of the internal registers. This method works fine until a fault in your program overwrites all of the RAM with garbage

and you have to reload. Unfortunately this happens many times when you are debugging a program.

The operating system

The latest generation machines overcome this problem by including a ROM which comes with the computer. This ROM contains a small operating system that continually controls the operation of the computer when your user program is not running. It is very similar to the monitor program found in an evaluation kit but it does not require a terminal to operate.

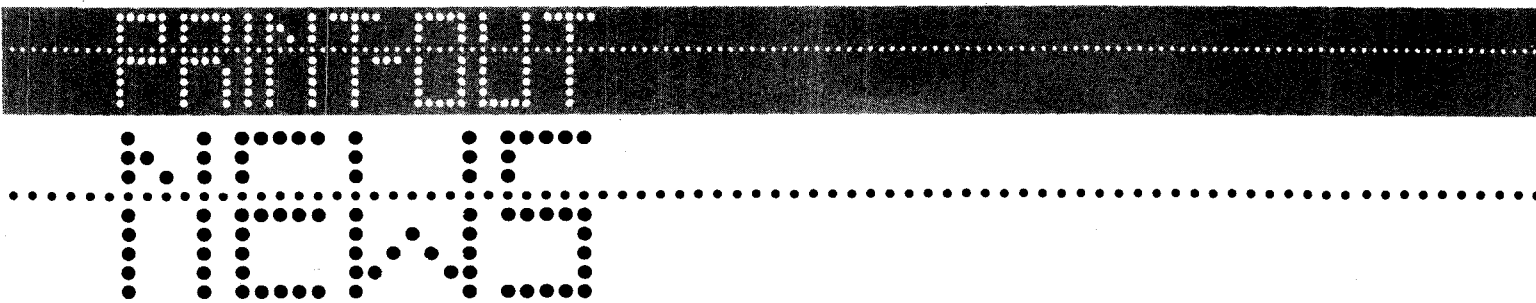
Instead it scans a low-cost calculator-style keyboard and LED display. The combination of the ROM's operating system and the cheap keyboard can provide all the functions offered with the evaluation boards and sometimes more, but at a fraction of the cost since a terminal isn't needed. Also included in the ROM are routines that allow easy examination of the internal registers' contents.

Since the program in ROM is permanent, these special routines are protected against corruption by a run-away program (your program gone wrong) and are also immediately available when the power is turned on.

Everything you wanted to know about hex . . .

With this operating system the latest generation computers also allow data manipulation in half-bytes of the older method of single-bit movement. In practice this means that the finger-sore operator loading his program has to press two switches, where before he had to press five. (In terms of 1000 or even 100 bytes that's a lot of finger presses). Not only is data entry simplified but also data output. Where earlier machines used a row of eight LEDs the present generation types display the byte as two hex characters.

That's where the hobby computer is today but who knows what it might become in a couple of years? Maybe a simple pen with a lead to plug in the back of your TV set?



TINY FLOPPY DISCS

At last manufacturers are doing something to overcome the big problem of expensive peripherals. Two American companies will soon be announcing tiny floppy disc drives using discs roughly 5" in diameter as opposed to the present 8". The discs are expected to hold something like 250k bytes and will be offered on the personal computer market at about half the present cost.

WHAT INTEL HAVE UP THEIR SLEEVE

With the announcement of the Z80 (see ETI News Digest last month) the days of the 8080 are numbered. Recently, however, Intel revealed that they have three new processors for the 8080 family. These are only a small part of an LSI product push scheduled for the next eighteen months.

The 8085 is a more powerful version of the 8080A, but twice as fast. Then there will be another processor which will be five times as powerful as the 8080A and Intel's first 16-bit processor which will beat the 8080A ten times over! As you can see these devices are aimed at the upper end of the microcomputer market.

For the low end of the market we are promised two minimum chip families: the MCS48 and the MCS41. The MCS48 family includes two processors, the 8048 and the 8748. Both of these contain, in addition to the CPU, ROM, RAM, and enough I/O capability for simple single-chip control applications. The chips in the family work with many of the 8080 peripheral chips. The 8748 contains a 8,192-bit field erasable ROM (it can be altered in the field with a UV light source).

RAM CUTS

RCA America is reported to be about to drop the price of their SOS RAMS by around 50%. These are the RAMs which are offered as part of the 1802 CMOS microprocessor system. They include the 256 x 4 & 1024 x 1 bit devices.

BRITISH BIPOLAR PROCESSOR

Ferranti of Great Britain have announced the release of the F100-L, a 16-bit single-chip microprocessor. Unlike the majority of other microprocessors, the F100-L is a bipolar device. Built on a 5.8mm square chip it contains approximately 7000 components and includes some six feet of aluminium interconnecting track!

Ferranti claim that the F100-L is the first microprocessor to be wholly developed, designed and manufactured in Europe. Released in conjunction with the device are the first two of a range of associated support chips, the F111-L control interface chip and the F112-L Data Interface chip for software support there is development software and a library of application programs.

COMPUTER CHAIN STORES

Riding on the growth of the computer hobby in America have come computer shops. These are retail shops where the potential customer can go to look at and actually have demonstrated (before buying) the different brands of computers and their support equipment.

So successful were the early stores that one man has set up a chain of computer shops across the USA. Paul Terrell, an ex-IBM engineer, opened his first 'Byte' shop in Mountain View, California in December 1975. Now, less than 12 months later, there are 17 Byte shops.

One report puts the number of computer shops in the United

States at over 50 and growing. Before they came along, hobbyists were forced to buy via mail order. Delivery was often slow and when equipment did arrive the purchasers had no-one to turn to for help if they had problems.

Here in Australia mail-order remains the only way to get a wide range of products, but things are improving as the electronics hobby shops start to catch on to the new trend. Many shops now sell evaluation kits; and Applied Technology seems to be well into things with a new computer Showroom at Hornsby.

AWAITING THE HOBBYISTS' CONTRIBUTION

"... we have every reason to foresee some really interesting and important technical developments coming out of the computing hobbyist activity" — July 1976 issue of the American IEEE journal, 'Computer'. Not only has the hobby side of computing gained acceptance by the professional, but it seems that hobbyists are being actively courted by the professional institution. And who can blame them? All those brainy people working away in their leisure time must be capable of developing all sorts of goodies that even the wealthiest of computer companies would not be able to afford if they had to pay for the labour!

A CASE FOR A MICROCOMPUTER

Microcomputers are finding their way into many strange places, but how about suitcases? Now offered is a microcomputer complete with microperipherals that comes in an attache style case. The peripherals include a 96-character keyboard, a miniature cassette-tape system and a 20-column alphanumeric display.

ALPHANUMERIC LED

Compact and complete with on-board electronics, HP's new LED alphanumeric displays are offered in standard DIPs, four characters in each package; they are end-stackable. On-board are shift registers and externally-programmable constant-current drivers, so associated circuitry is simplified. Compared with the circuitry required for earlier and more costly alphanumeric LED displays, the parts count for a typical 32 character system has been reduced by a factor of 36 to 1. It takes only 12 pins to address each four-character set. These low-voltage LED displays are directly TTL-compatible and readily microprocessor-controlled. Each HDSP-2000 character is formed with a 5 x 7 dot matrix, capable of displaying the full ASCII font, lower-case as well as upper-case letters, punctuation marks, mathematical and other symbols, as well as numerals. Each four character package measures 17.7mm (0.699 in.) long by 7.25mm (0.290 in.) high.

For further information contact Amphenol Tyree.

NSW: Ray Gudgeon, 176 Botany Street, Waterloo, 2017. Telephone: (02) 69 5264.

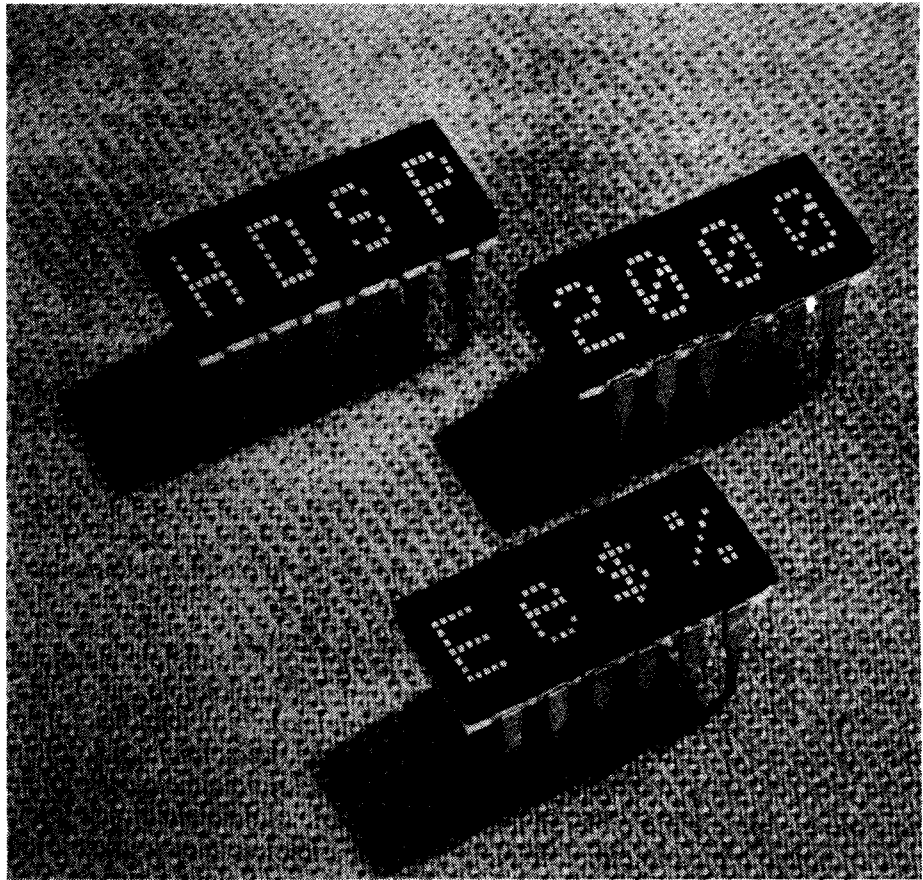
VIC: Richard Knehans, 115 Highbury Road, Burwood, 3125. Telephone: (03) 288 7099

6800 COMPILER

Motorola Semiconductors have announced that they will soon offer a resident Fortran compiler for use on the 6800 Exorciser development system. The compiler requires 16 kilobytes of memory and is close to the ANSI standard.

THREE NEW LOCMOS MEMORIES

Philips Electronic Components and Materials announce the release of three new 256 x 1 random access memories as part of their



The Hewlett-Packard HDSP 2000 alphanumeric LED displays.

range of LOCMOS integrated circuits. These are the HEF4720, HEF4720V and HEF4720V/S1.

The HEF4720 and 4720V are static RAMs with a stand-by power consumption of typically 5 μ A. They are specially suitable for memory systems in battery-operated equipment. The HEF4720 has a supply voltage range from 3 to 15 V and an access time at 5V of 550 ns.

MICROS MOVE INTO AUTOMOBILES

The US auto giant General Motors is expected soon to name the prime source of the microprocessor it will use in its cars of the future. Four companies are apparently on the short list Intel, Motorola, National Semiconductor and Texas Instruments.

NEXT MONTH'S PRINTOUT

- What is a 'systems approach' and why should you know about it before you buy or build your computer?
- How should you choose which microprocessor to base your system on?
- What is an operating system and how should this affect your choice of computer?
- Plus other advice, news, and features.

ETI PROJECT 804 SELECTA TV GAME

— play 6 different games with on-screen digital scoring and sound effects.

As described in November 1976 Electronics Today, this must be one of the most popular TV game projects yet published. Designed around the AY-3-8500 IC by General Instruments this single control chip offers a choice of six games together with ON-SCREEN scoring and sound effects.

Although not a project for absolute beginners the kit can be assembled by anyone with reasonable experience in electronics.

The HOBBY KIT SELECTA TV GAME comes complete including all parts as listed in the article, full instructions and pre-wound coils for the modulator circuit.

NOTE: AS A SPECIAL SERVICE WE HAVE PRETESTED EACH AY-3-8500 IC FULLY BEFORE DISPATCH.

We guarantee that each device is fully operational when it leaves our factory.

PRICE WHILE STOCKS LAST \$49.50
+ \$3.50 pack, post and insurance.

PLEASE NOTE:

We have a large but limited quantity of these kits available. All orders will be promptly despatched in strict priority of receipt. Naturally any order received after stocks are depleted will be refunded immediately. Sorry, no phone orders or C.O.D. please.

AUDIO PRODUCTS

Our HOBBY KIT audio products have proven very popular with hobbyists everywhere. You can enjoy building these exciting projects and save a fortune over the cost of equivalent mass produced products.

For full details write for our free catalogue.

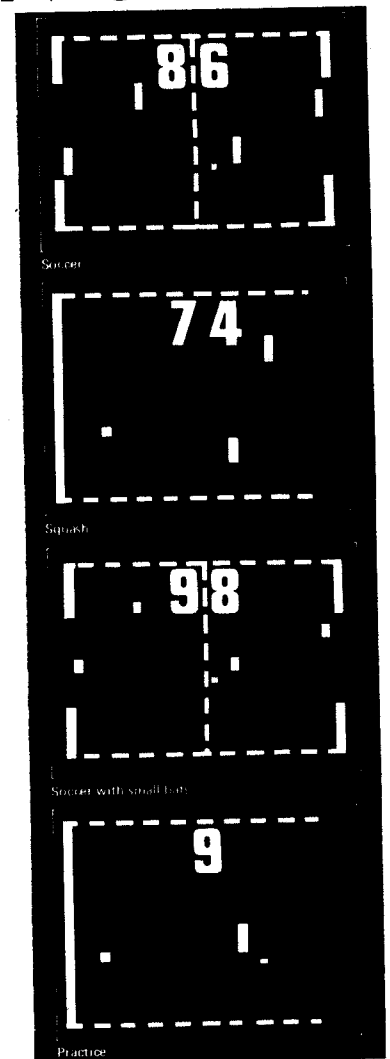
- TWIN TWENTY FIVE 25W RMS AMPLIFIER \$85.00
- TWIN FORTY 40W RMS AMPLIFIER \$99.50
- ETI 740 VARICAP FM TUNER \$120.00
- AD8K40 40W SPEAKER KIT (PR) \$125.00
- AT1265 30W 3 WAY WRAP AROUND SPEAKERS (PR) \$195.00

SPECIAL OFFER (Closes December 31, 1976)

STR100 PROFESSIONAL STEREO TEST RECORD

Evaluate your Hi Fi system with this exceptional record. Included with the record is a 16 page booklet by Audio's Edward Tatnall Canby explaining how to use the record to improve the performance of your system.

Catalogue price \$12.50 but if purchased with one of the above kits \$6.25.



POPULAR HOBBY KITS

MICROPROCESSORS

We still have stocks of the incredible SC/MP INTRO KIT and have developed a number of peripheral kits to use with the SC/MP. Our extensive range of microprocessor components has now been enlarged with LOW POWER SCHOTTKY TTL (see next page) as well as PROMS, RAMS, UARTS, CHARACTER GENERATORS AND VOLTAGE REGULATORS.

- SC/MP INTROKIT \$89.50
- AT1250 SC/MP POWER SUPPLY \$37.50
- PC2102/8 1K x 8 RAM KIT \$32.50
- PC900 PROTOTYPE PCB \$9.50



HOBBY NEWS

NOVEMBER, 1976

AT LAST! LOW POWER SCHOTTKY TTL

At last we have been able to secure a comprehensive stock range of this exciting new logic family. These devices require no special handling as with CMOS, they are considerably faster (45MHz typical) and consume about one fifth the power of equivalent TTL functions. You can use these devices to upgrade conventional 74 series designs as they have identical pin connections and in most cases will directly interface with other TTL families without the need for buffer circuits. We have full data sheets and applications information available if required (20c/division).

* All prices include postage/packaging and sales tax.

74LS00	QUAD 2 INPUT NAND GATE	.39
74LS01	QUAD 2 INPUT NAND GATE (OC)	.39
74LS02	QUAD 2 INPUT NOR GATE	.39
74LS03	QUAD 2 INPUT NOR GATE (OC)	.39
74LS04	HEX INVERTER	.43
74LS05	HEX INVERTER (OC)	.43
74LS08	QUAD 2 INPUT AND GATE	.39
74LS09	QUAD 2 INPUT AND GATE (OC)	.39
74LS10	TRIPLE 3 INPUT NAND GATE	.39
74LS11	TRIPLE 3 INPUT AND GATE	.39
74LS13	DUAL NAND SCHMITT TRIGGER	.86
74LS14	HEX SCHMITT TRIGGER	2.02
74LS20	DUAL 4 INPUT NAND GATE	.39
74LS21	DUAL 4 INPUT AND GATE	.39
74LS27	TRIPLE 3 INPUT NOR GATE	.43
74LS28	QUAD 2 INPUT BUFFER	.43
74LS30	8 INPUT NAND GATE	.39
74LS32	QUAD 2 INPUT OR GATE	.47
74LS37	QUAD 2 INPUT NAND BUFFER	.47
74LS38	QUAD 2 INPUT NAND BUFFER (OC)	.47
74LS40	DUAL 4 INPUT NAND BUFFER	.43
74LS42	BCD TO DECIMER DECODER	1.53
74LS73	DUAL J-K MASTER SLAVE F/F	.51
74LS74	DUAL D TYPE EDGE TRIGGERED F/F	.61
74LS75	QUAD BISTABLE LATCH	.86
74LS78	DUAL J-K.NEG EDGE F/F	.51
74LS86	QUAD 2 INPUT EXCLUSIVE OR GATE	.67
74LS90	DECADE COUNTER	1.37
74LS92	DIVIDE BY 12 COUNTER	1.37
74LS93	4 BIT BINARY COUNTER	1.37
74LS95	4 BIT L R SHIFT REGISTER	1.84
74LS109	DUAL J-K POS EDGE F/F	.61
74LS113	DUAL J-K POS EDGE F/F	.61
74LS114	DUAL J-K NEG EDGE F/F	.61
74LS151	8 TO 1 MUX	1.80
74LS157	QUAD 2 IN DATA SELECTOR (NI)	1.71
74LS163	SYNC 4 BIT BINARY COUNTER	2.86
74LS164	8 BIT P OUT SERIAL S R	2.02
74LS174	HEX D-TYPE F/F WITH CLEAR	1.84
74LS175	QUAD D-TYPE EDGE TRIGGERED F/F	1.84
74LS181	4 BIT ARITHMETIC LOGIC UNIT	4.59
74LS191	SYNC BINARY UP/DOWN COUNTER	3.06
74LS192	SYNC DECADE UP/DOWN COUNTER	3.06
74LS193	SYNC 4 BIT UP/DOWN COUNTER	3.06
74LS194	4 BIT BIDIRECTIONAL S/R	1.84
74LS195	4 BIT PARALLEL S/R	1.84
74LS196	DECADE COUNTER LATCH WITH PRESET	1.84
74LS221	DUAL MONOSTABLE MULTI	1.74
74LS253	DUAL 4 TO 1 DATA SELECTOR/MUX	1.88

PROJECT KITS

PROJECT KITS – now everyone can enjoy building solid state Electronic Kits.

This popular range of kits appeals to everyone. Each PROJECT KIT is supplied with a predrilled fibre glass printed circuit board, all components including solder and hookup wire and simplified assembly instructions. All prices quoted include postage and packing.

PK061	Simple Amplifier	\$6.00
PK044	Two Tone Door Bell	\$5.00
PK043	Coin Tossor	\$3.50
PK068	Led Dice	\$6.00
PK1000	General Purpose IC Amplifier	\$5.00
PK2000	Variable Timer	\$7.50
PK3000	Touch Switch	\$7.75
PK4000	Burglar Alarm	\$10.95
PK5000	3 DIGIT Readout Module	\$19.50
PK5050	Digital Voltmeter Module	\$8.50
PK5080	Frequency Meter Module	\$11.50
PK7000	Spot-O-Light Display	\$9.50
PK8000	Bar-O-Light Display	\$9.50
PK9000	Digital Clock Module	\$13.50
	(transformer for PK9000)	\$7.50

For full details please refer to our catalogue in ETI August 1976.

FOOTNOTE:

Our August Catalogue has brought us literally thousands of new customers. To these people and especially our existing customers we would like to say "many thanks" for your support.

To cope with the enormous increase in orders our mail order department have put in many hours of overtime and at last they have restored things to a more reasonable situation.

It appears that so many people have given us their business that our major competitor has become almost desperate. Please give him back at least some of your support as this will enable us to restore our grade of service to our pre-catalogue level.

Many thanks once again.



POSTAL ADDRESS: THE ELECTRONIC MAILBOX, P.O. BOX 355, HORNSBY 2077.

WAREHOUSE & SHOWROOM: 109-111 HUNTER STREET, HORNSBY, N.S.W. 2077.
Telephone: 476 3759 476 4758

TRADING HOURS: 9-5 WEEKDAYS, 9-12.30 SATURDAY

MICROCOMPUTER TERMINAL -PART TWO

Last month we announced our plans for a microcomputer terminal project. Constructional details will start next month but in the meantime here are details of some of the components and techniques used...

LAST MONTH WE ANNOUNCED our plans for a computer terminal project; we showed how the terminal would be in three main parts so you do not need to build the whole thing if you already have a keyboard or a printer, or if you want parallel interfacing rather than the teletype interface we are proposing.

We will start to publish constructional details of this project next month, but now we'll give you an idea of the basic concepts behind the design.

We want this part of our terminal to

The Keyboard

give out the codes you get when you use a commercial teletype machine.

These codes appear on an eight-line data bus and they correspond to the codes given in Figure 1. There is another output from the keyboard which is used to tell the computer when to read the data on the bus.

From Figure 1 you will see that there are more symbols on the computer keyboard than you find on the

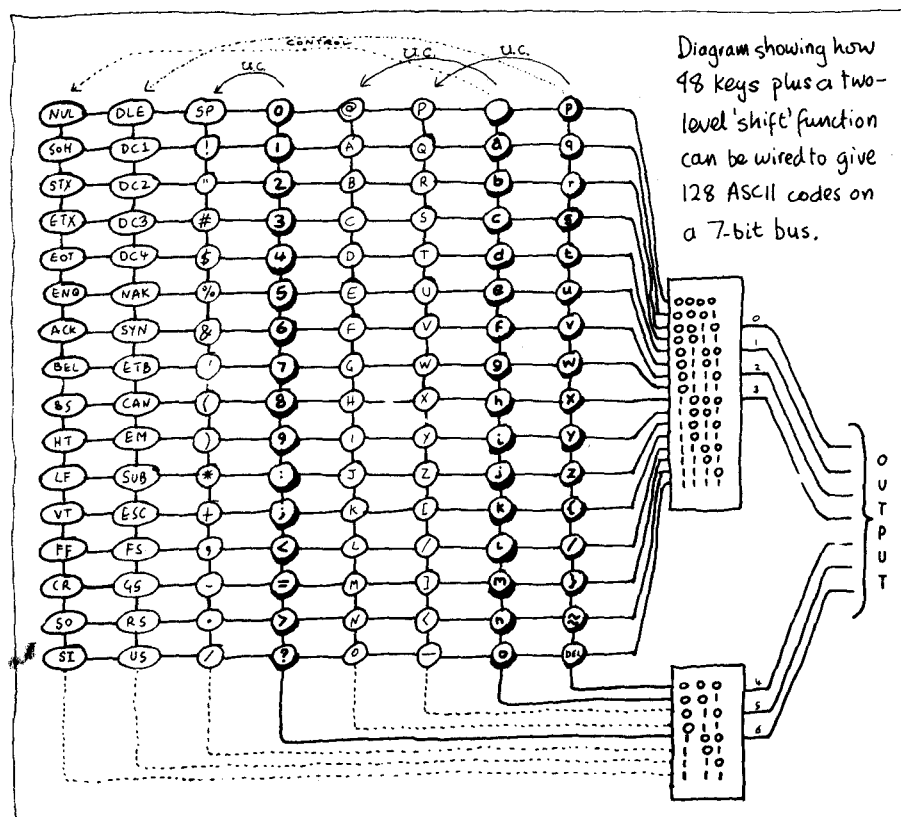
common typewriter. Apart from the usual alpha characters in upper and lower case and the numeric characters and their corresponding upper case symbols, there is a further shift which enables control codes to be entered. These codes will be explained later.

Figure 1 also shows that there is a maximum of 128 codes, and that these are put out on seven parallel lines. The eighth line of the bus carries the parity bit. This bit is high or low according to the status of the other seven bits and is used as a check against errors.

The TV display part of the computer is only able to display 64 characters — upper case alphas plus the numerics and a few symbols. So that these characters can be inputted from the keyboard without having to operate the shift control we will incorporate a special switch on the keyboard to give 'upper case only' on alpha characters.

The electronics will centre around a special keyboard encoder IC which has a 16-line input and a 4-line output. The outputs range from 0000 to 1111 when the inputs are activated one at a time. This IC gives four of the lines for the output bus; the others are taken from logic which looks which part of the keyboard you are using and whether or not you have operated the shift key.

The connections to the switches on the keyboard will have to be made using wire links — we considered designing a double-sided pcb but we thought that the cost of the board was too high to make it worthwhile.



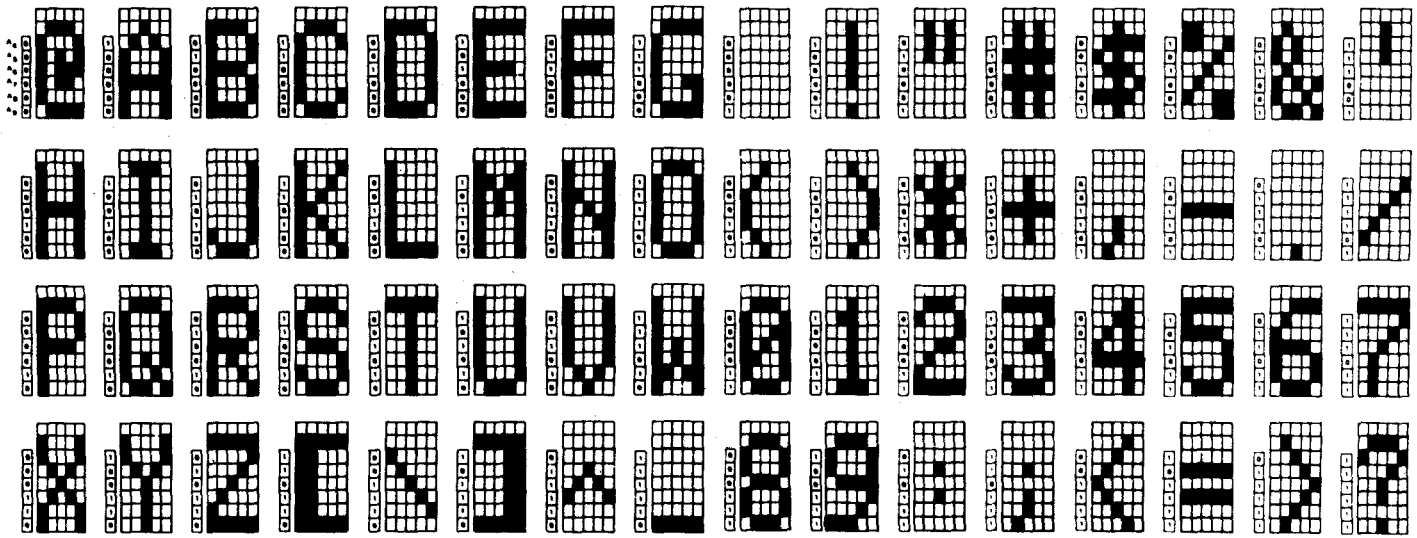


Fig.2. This table shows the font of 64 characters that the 2513 IC can put onto a TV screen. To the left of each 5x7 matrix the six-bit code which specifies the character is given. This code is derived from the seven-bit ASCII code from the keyboard.

The VDU

This part of the terminal takes in data on a parallel bus and stores up to 256 bytes in RAM. Then it decodes the characters and generates the video necessary for them to be displayed on a TV screen.

The electronics centres on the Signetics 2513 character generator IC. Figure 2 shows the characters as they appear on the TV screen when any of the 64 possible codes (only 6 lines of the bus specify characters) are inputted. The characters are made up of dots on a 5 x 7 matrix — the device has to specify five bits of information at exactly the right place on seven consecutive TV lines. The eighth line has to have a blank space to give spacing between lines of text.

Figure 3 shows how each of these eight lines is given an address from 000 to 111 and it shows how the five bits of data leave the IC in parallel. In all the IC has nine pins for inputting data and five pins for outputting data. The first three input pins carry the row address (the line address) and the other six specify the character.

Figure 4 shows that the project we are designing includes counters to set up the row address. There are counters too, to find the right address in the RAM so that the data fed into the 2513 changes as the electron beam scans across and down the screen.

The parallel data coming out of the

character generator has to be serialised and made into a proper video signal before the TV can handle it. This means it has to have line and frame sync pulses inserted and the timing of the data which draws the characters on the screen has to be precisely fixed in relation to these pulses. A crystal oscillator will be used to keep everything in step.

If you intend to build this project and want to get on with something useful you can start by converting your TV set into a video monitor. We recommend this approach rather than trying to use a modulator — if you can't modify your TV now wait until next month and we'll tell you how to do it.

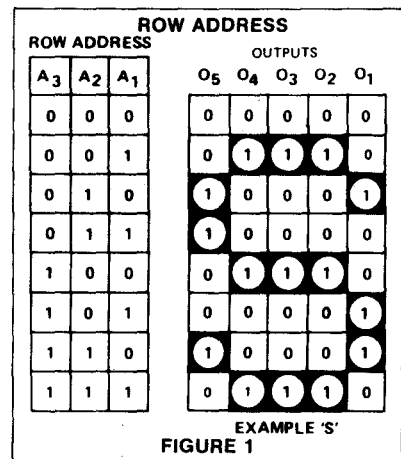
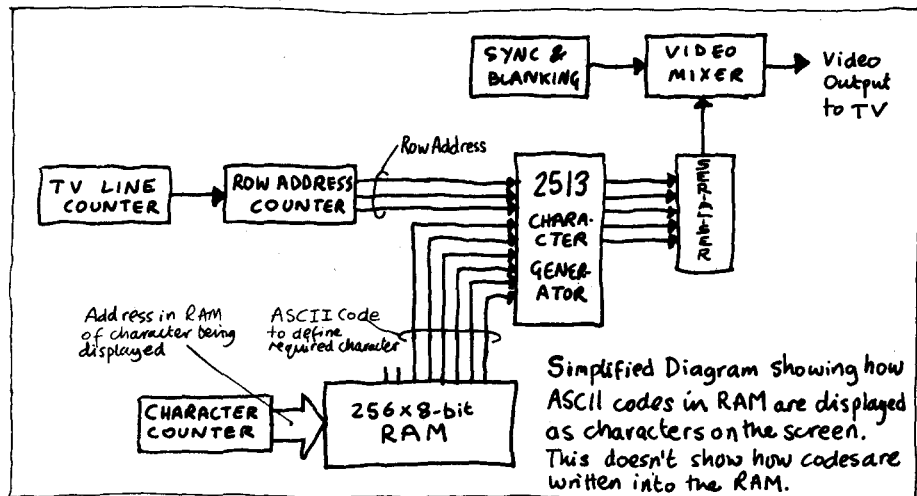
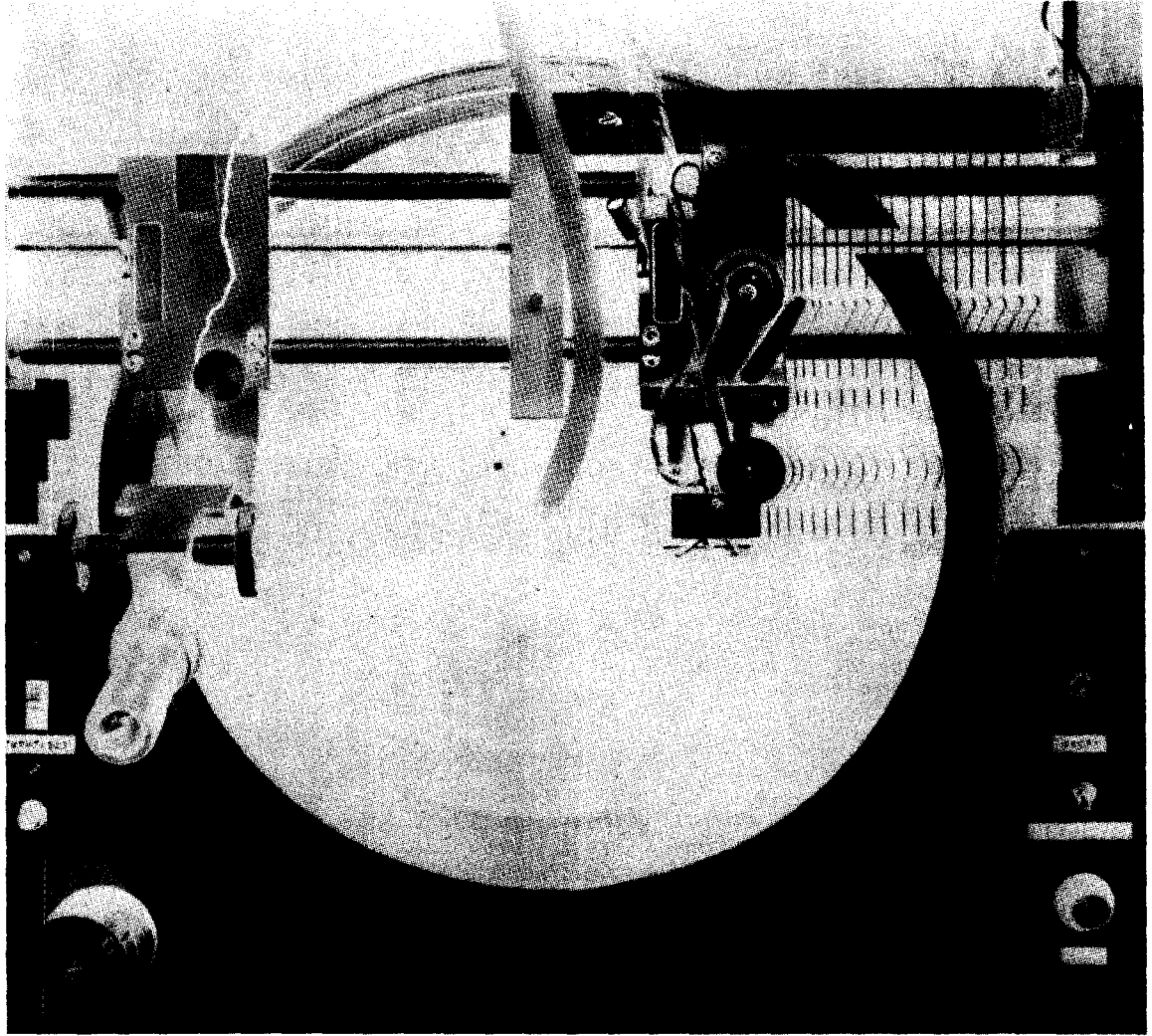


Fig 3. To put a character on the screen the character generator outputs seven rows of data on its five output pins.



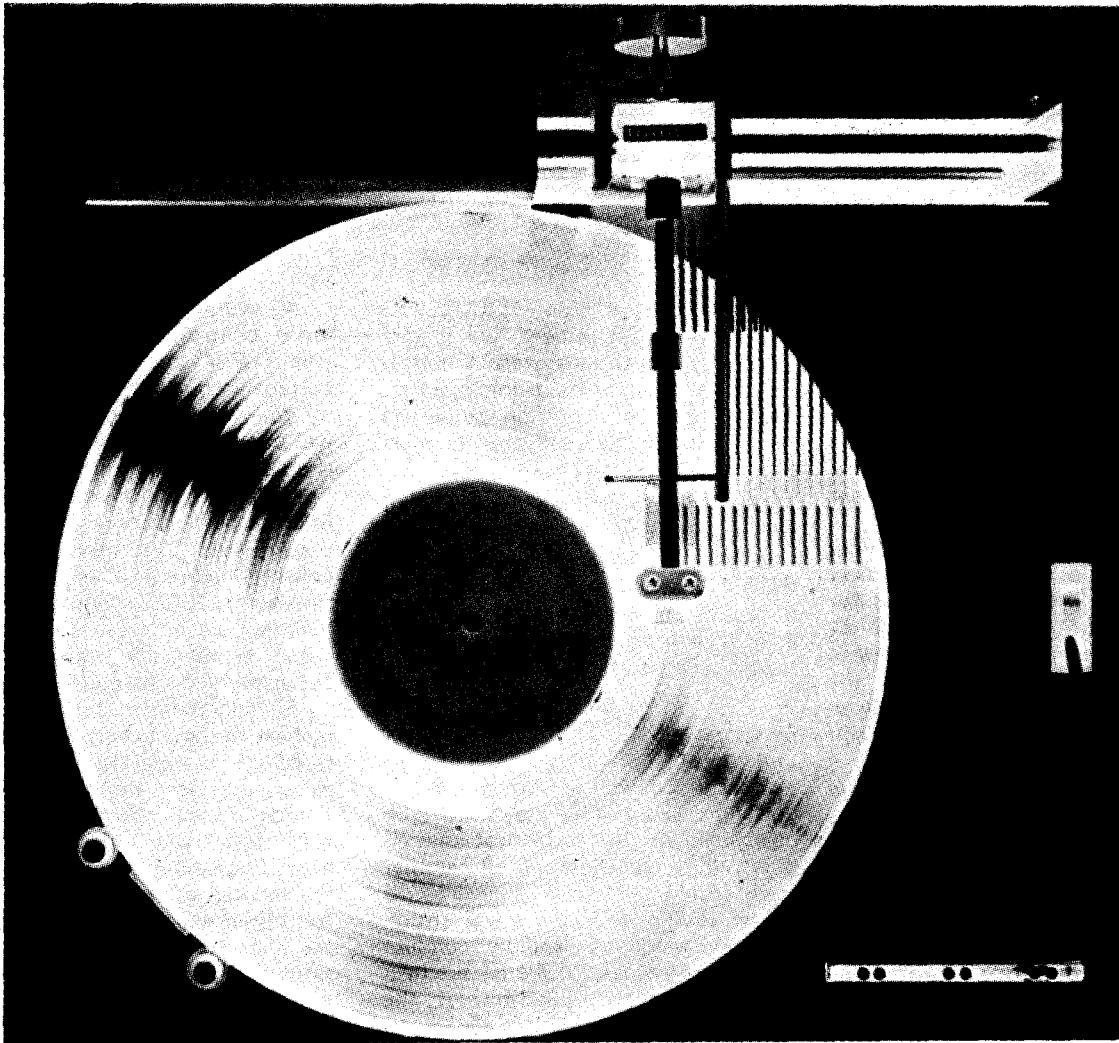


This is how discs are made.

A master disc is cut on a special lathe. The cutting head moves across the master in a straight line from the edge to the center. The special stylus inscribes a groove in the surface of the disc.

Ideally, a turntable system should enable the stylus in your cartridge to meticulously follow the "path" inscribed during the cutting process. That is, it should play your record precisely as the master disc was originally cut.

A "straight line tracking" turntable system, properly designed, engineered and manufactured, could eliminate problems such as skating force, tracking error, and the resulting excessive record wear, all of which are *inherent* in pivoted arm systems in all their forms and modifications.



This is how the ST-7 plays them.

The Rabco ST-7 is a *straight line tracking turntable*. Your stylus precisely follows the original path cut into the master record. The result is the total elimination of *both* tracking error and skating force.

The ST-7 begins with straight line tracking. In every other respect—motor, suspension, bearings, drive, controls—it is exemplary of a professional instrument designed for home use.

The ST-7 offers a cascade of zeroes. Zero tracking error. Zero skating force. Zero stylus overhang. Zero horizontal friction. Zero vertical friction.

The ST-7 plays music in the home in a way that makes conventional pivoted arm systems obsolete.

For complete information write to Harman Kardon, P.O. Box 6, Brookvale, 2100.

harman/kardon

This is the 36th and last part of this series.

The intention has been to provide introductory information about modern electronics for intelligent people who had no prior training in the subject. Surveys conducted throughout the series have established that we have achieved this aim.

The author wishes to thank Brian Chapman, formerly Technical Editor of Electronics Today International for his patient and conscientious job of editing the whole series. Maree Breen has made a wonderful job of the author's rough sketches; Colleen Drummond typed the first manuscript drafts, Jan Vernon turned the original monthly articles into the final three-volume form.

Finally, Electronics Today International wishes to give its heartiest thanks to the very many companies who provided information and illustrations used. We are most grateful for their willing and prompt cooperation.

Peter Sydenham,
Armidale, October, 1976

IN PART 10, WHEN DISCUSSING the types of amplifiers, we briefly mentioned the power stage found at the output end of electronic systems. Typical devices requiring amplifiers to drive them are loudspeakers, electric motors, and heaters.

The power handling capability of the various designs of these special amplifiers can range from one watt to many kilowatts. In this final part we introduce the special semiconductors and techniques used in electronic power control.

HEATSINKS

As some power is lost as heat in power transistors they may usually be recognized by the large heatsinks on which they are mounted. A rectifier stage using flat-plate heatsinks is shown in Fig. 1. These metal structures are needed to rapidly conduct away and dissipate to the air the heat generated at the junction of the device — this is a critical design requirement. The approach to designing heatsinks is common to all power components.

All semiconductors used in analogue control will have heat losses (the power lost as heat equals the current through the device multiplied by the voltage drop across it) which will cause the junction temperature to rise above the case outer temperature. For example, a transistor power amplifier stage may have at half output power (say) 10 V drop and 10 amp collector current. The heat loss is, therefore, 100 W and this must be liberated in order to keep the transistor temperature lower than its recommended maximum value.

All materials resist the conduction of heat to some extent — this property is called 'thermal resistance' and its value depends upon the material (copper is less resistive to heat flow than iron) and the cross-sectional area (increasing the area decreases the resistance). In practice catalogues for power components usually quote the thermal resistivity θ (which has units $^{\circ}\text{C}/\text{W}$) between two points on the device. For example, typical measured temperatures for a certain power transistor mounted on a heat-sink are as shown in Fig. 2. From these temperatures we can see that: —

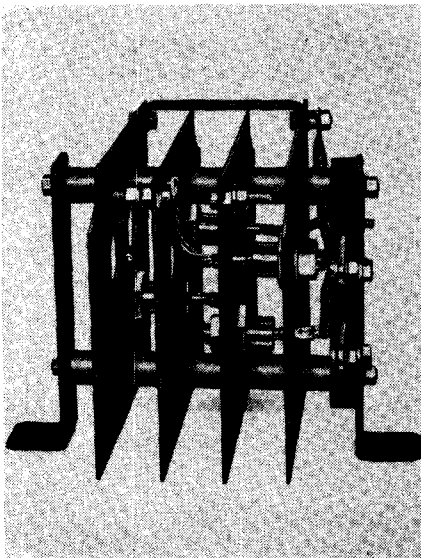


Fig. 1. Power handling stages are easily identified by the heatsink assemblies on which they are mounted — a 35 A, 60 VDC rectifier stack is shown here.

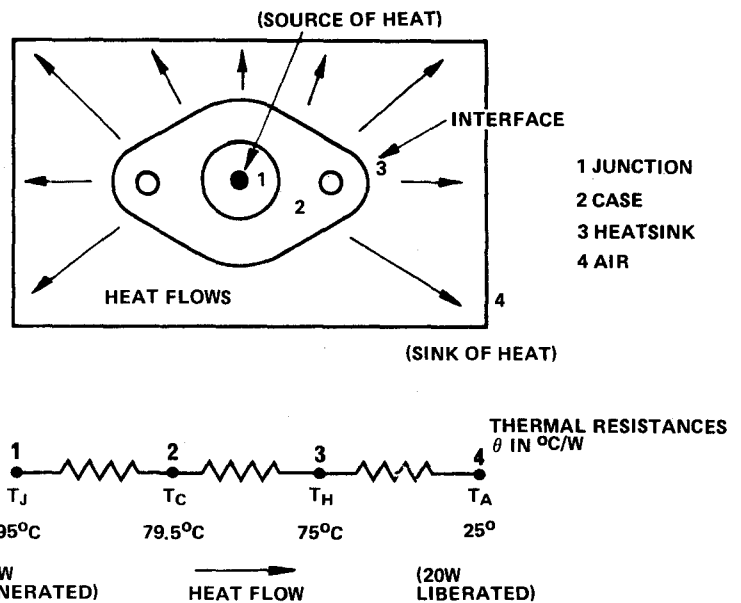


Fig. 2. Thermal resistances for a 20 W power transistor.

TABLE 1

$$\theta_{J-C} = (95 - 79.5)/20 = 0.77 \text{ }^\circ\text{C/W}$$

$$\theta_{C-H} = (79.5 - 75)/20 = 0.23 \text{ }^\circ\text{C/W}$$

$$\theta_{H-A} = (75 - 25)/20 = 2.5 \text{ }^\circ\text{C/W}$$

$$\theta_{J-A} = (\theta_{J-C} + \theta_{C-H} + \theta_{H-A}) = 3.5 \text{ }^\circ\text{C/W}$$

Where J = junction, C = case of device, H = heatsink and A = air.

From this example we can see that the thermal resistance within the device — the parameter the user has no control over — is larger than the case-to-heatsink value. This means it is not worth improving the contact and heatsink material. The important thermal resistance is that between the junction and the air (presumed to be at constant ambient value); in many cases a different shape heatsink, one that transfers heat better to the air (finned for example) would make an improvement. The thermal resistivity (heatsink to air) can also be reduced by forcing air past the heatsink and/or by increasing the heatsink surface area. The latter measure, however, also has its limits because the thermal resistance between the device connection point and extremities of larger plates rises with increasing dimensions (reducing the effectiveness of outer areas).

The above example illustrates how a heatsink stage can be designed using the concept of series thermal resistances. In practice the design procedure must be worked in reverse. The aim is to ensure that the junction temper-

Material used between device and heat sink (for insulation)	Thermal Resistance θ_{C-H} in $^\circ\text{C/W}$	
	Dry	with heat conducting grease
Direct contact (TO3)	0.20	0.10
Teflon insulator shim (TO3)	1.45	0.80
Mica shim (TO3)	0.80	0.40
Anodized aluminium (TO3)	0.40	0.35
0.25in stud mount (direct)	0.40	0.25
0.50in stud mount (direct)	0.12	0.07
0.75in stud mount (direct)	0.07	0.04

Fig. 3. Table of thermal resistances θ_{C-H} for typical mounting methods. Values can vary widely.

ature remains less than a specified maximum limit. Beyond this quoted value the junction will be destroyed. A practical difficulty is that the junction temperature cannot be measured to ensure that the design is adequate so selection of mounting and heatsink type must be made with care using manufacturers' quoted thermal resistance values as the basis of a design. The following steps are given as a guide but full detail should be sought from more detailed accounts — see reading list.

Step 1: Assess the maximum power (W_{max}) to be dissipated by the device. This will be the worst case of V.I product remembering to allow for temperature effects and maximum

values. In switching designs the base to emitter junction voltage of a transistor is significant.

Step 2: Establish T_{Jmax} , T_{Amax} from data sheets and expected ambient conditions. This enables the minimum required value of T_{J-A} to be calculated.

Step 3: Calculate the overall thermal resistivity needed from $\theta_{J-A} = T_{J-A}/W_{max}$.

Step 4: Establish θ_{J-C} and θ_{C-H} from device table charts and the mount thermal resistivity for the device clamping method. Fig. 3 lists typical θ values for various clamping methods.

Step 5: Calculate θ_{H-A} required

$$\theta_{H-A} = \theta_{J-A} - (\theta_{J-C} + \theta_{C-H})$$

Step 6: Use heatsink tables to find

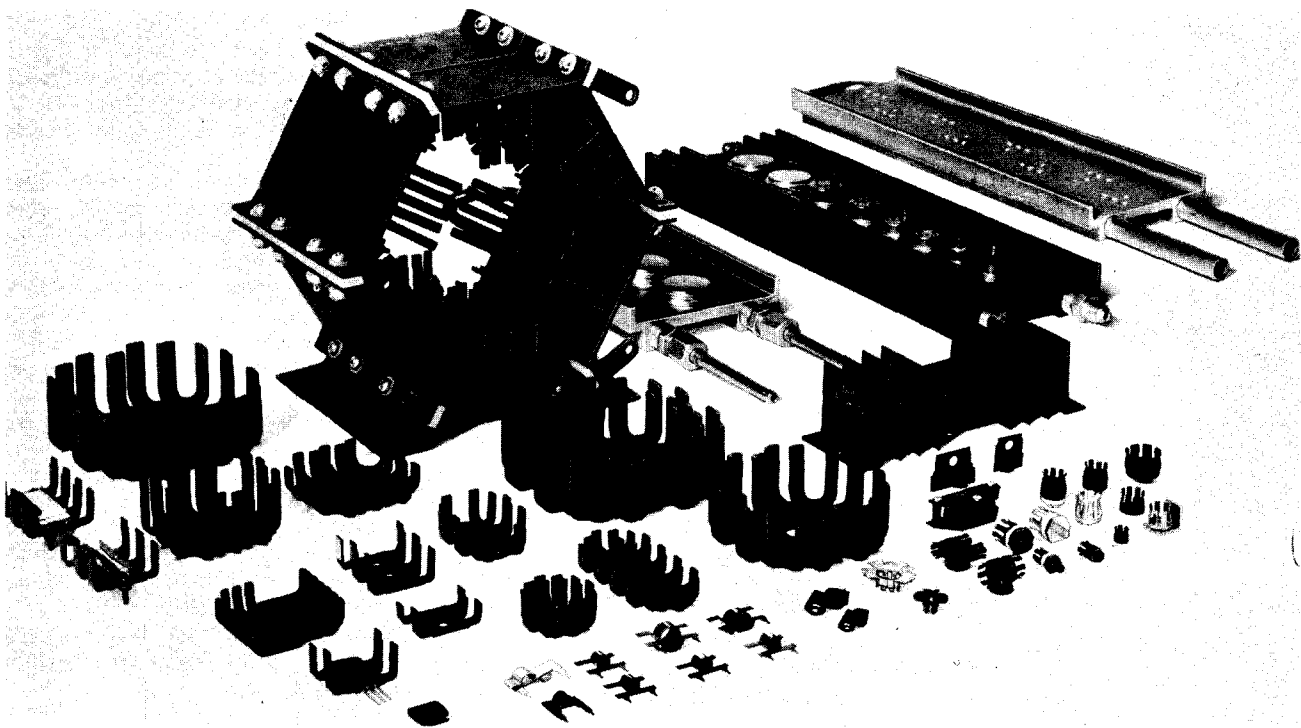


Fig. 4. Range of heatsinks for dissipating excess heat in semiconductors over a range from milliwatts to kilowatts.

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suitable design having θ_{H-A} value or smaller.

In general if θ_{H-A} needs to be less than 2 to 5 °C/W the heatsink becomes prohibitively bulky. Design of the whole system is usually limited by the manufacturer's value of θ_{J-C} , which cannot be reduced. The interface coefficient θ_{C-H} is usually around 0.15–0.20 °C/W for direct contact using the recommended heat conducting silicon grease. Mica insulation degrades this value a little, poor heat conducting insulators should be avoided as they contribute a quite high value of θ_{C-H} .

Heatsinks for analogue control power units will need to be much larger than those of switching designs such as the switching regulator and normal rectifier stacks. This is because the latter need only dissipate the V.I product of the two extremes of V and I. The voltage drop across a power diode running at many amperes is around one volt: when reverse biased the voltage is high but the current negligible.

Figure 4 shows a wide selection of heatsinks including units for fluid cooling applications. Fins should always be positioned to assist the vertical convective flow of air over the surfaces. Total immersion of the electronic circuit in cooling liquid is not used.

POWER TRANSISTORS

Power transistors are little different to small-signal devices in their basic semiconductor principle of operation: the distinguishing factors are the heavy-duty design which enables high collector currents and voltages to be controlled. The junction areas are much larger and the case design is made to keep the thermal resistivity as low as possible (around 0.8 °C/W) in order that the losses can be removed. Collector currents being higher and the gains being lower than small-power transistors means the base currents are also large. Thus, high power stages have to have lesser power stages driving them. They are available

for several hundred volts operation and current levels exceeding a 1000 A. Cut-off frequencies into the gigahertz region are available (with less gain than that of lower frequencies). At RF frequencies gains range from 4-13 dB for powers in the range 0.1–80 W. There are few power applications that transistor devices cannot handle. In practice, however, certain other semiconductor devices are often a better choice.

SCRs, THYRISTORS AND TRIACS

Semiconductors and diodes have one p-n junction and transistors have two junctions, p-n-p or n-p-n. A logical progression is the three-junction device, p-n-p-n. This family contains such devices as the silicon-controlled rectifier SCR, the silicon-controlled switch SCS, the gate-turn-off switch GTO, the light-activated, silicon-controlled switch LASCs, and the Shockley diode. Of these, the SCR (also called a thyristor) mainly concerns us as it is able to control high-power levels (they were introduced in Part 16). The SCR has an anode and cathode and a gate lead (which when held positive prevents the unit from conducting).

By controlling the gate voltage it is possible to control when power begins to flow during an ac cycle. Once the SCR is triggered (or fired) it remains on until the anode-cathode voltage falls to zero again. SCRs are, therefore, extremely useful when an alternating current source is available as this automatically provides the necessary switch-off conditions at each half cycle.

TRIACS are special SCRs that can be switched on to allow both positive and negative half cycles to pass. This action can also be arranged by using two SCRs.

This class of device cannot control the flow of dc power from a dc source, because once turned on they remain on, acting like an adequately low-resistance contact. They are, however, invaluable for controlling loads which can be energised by ac power — heating coils, motors, lighting and furnaces.

The operating circuitry for an SCR is designed to provide the appropriate gate on-voltage level at the correct time during the half cycle. Fig. 5 shows five basic forms of phase control. A typical trigger circuit is given in Fig. 6. One difficulty in this kind of control is that large line transients are generated, along with RF interference, when the power begins to flow during each cycle.

A more refined type of control derives the required average output power as the mean of a series of complete whole-cycles rather than in

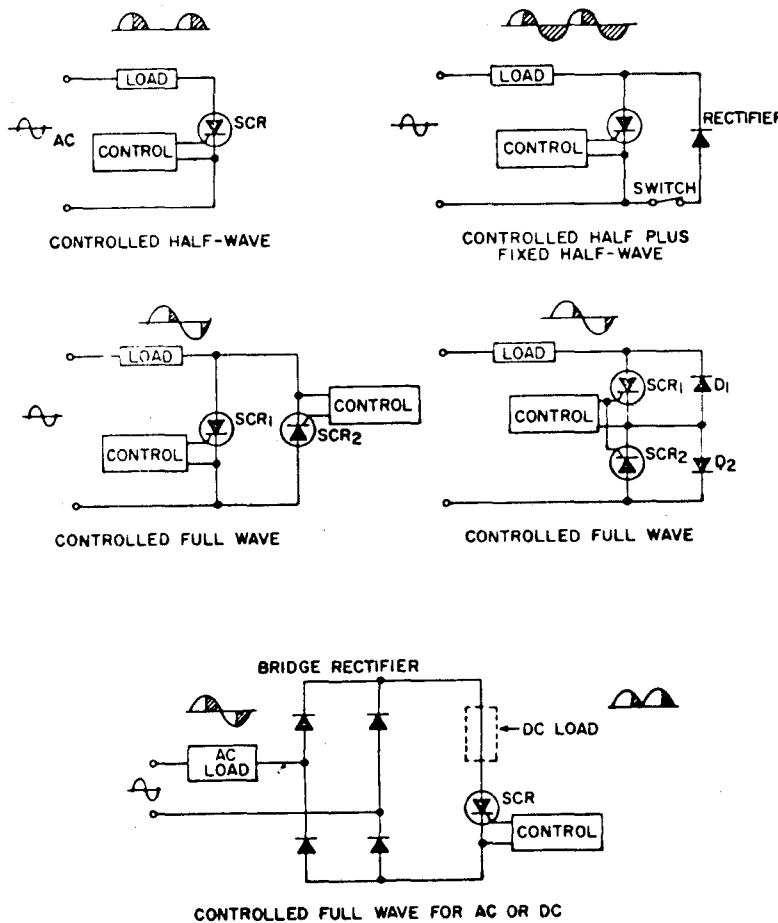


Fig. 5. Five arrangements by which a load can be fed with power flow controlled by SCR devices.

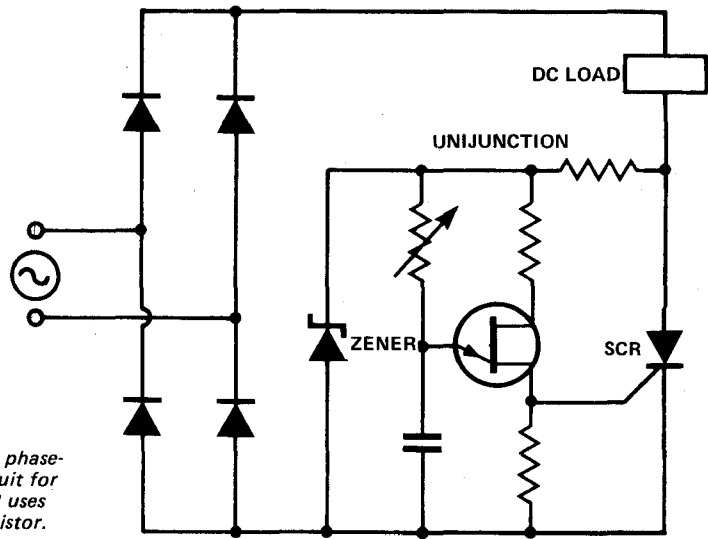
the mean of many partial cycles. This method generates substantially reduced line transients and RF interference because switching always occurs at the zero voltage condition: Figure 7a shows one form of proportional zero-voltage-switching controller using a TRIAC to control the heat produced in the element. Figure 7b is a typical output signal burst of gradually increasing power.

Capabilities of SCR devices range to hundreds of amperes, reverse voltages to as much as 2000 V. The maximum voltage drop across the turned-on SCR lies in the range 1.3–2.5 V, with leakage currents being in the region of 40 mA in the turned-off state.

These characteristics may make SCR devices appear extremely robust. Design of reliable, high-power, units, however, is a matter for a specialist. Many pitfalls can occur if their operation is not understood in detail. Designed properly they will, however, give utmost reliability.

Fuses for SCR circuitry also need special consideration because semiconductor junctions when overloaded will blow more rapidly than simple wire fuses or electromagnetic circuit breakers. The criterion is that the I^2t rating of the SCR must be greater than that of the fuse. I^2t values are usually provided in maker's data sheets. During the turn-on period of the SCR this value may drop significantly. Selection of adequate protection fuses is a matter that must be studied in some depth. Care must be taken to mend blown fuses in SCR units with the correct replacement — this invar-

Fig. 6. This typical phase-control trigger circuit for controlling an SCR uses a unijunction transistor.



ably means carrying the correct spare ready to use.

SWITCHING REGULATORS AND CONTROLLERS

Parts 15 and 16 discussed methods used to regulate dc power supply output voltages or currents. For small power levels — a few watts — the series regulator and zener diode arrangements are acceptable because the power they dissipate is not an economic factor. The controlling transistor (as is shown diagrammatically in Fig. 8) can, instead, be used as a switch varying the on-to-off time ratio (mark-space is the term used) in order to vary the average dc power obtained after smoothing.

The switching method has the

significant advantage of very small losses in the regulator stage. The transistor is either fully-on (high current but very low voltage) or fully-off (highest voltage but minimal current). As well as reducing the losses the method also can use a smaller capacity transistor. The price paid is the need for a filter stage and for a pulse generator to drive the switch.

Switching regulators are especially necessary when the voltage drop between the source and the load requirements becomes large.

Modern designs often make use of an integrated circuit as the basic control unit adding an additional switching transistor to cope with the output current needed. Fig. 9 is a high-current switching regulator which can supply 3 A continuously at 30 V input with losses sufficiently small to allow the use of quite small heat-sinks.

Switching is also a suitable method to efficiently control output loads — the difference between this and regulator design is that the feedback loop (dotted in Fig. 8) is not used; the mark-space ratio of the generator being controlled instead by the input

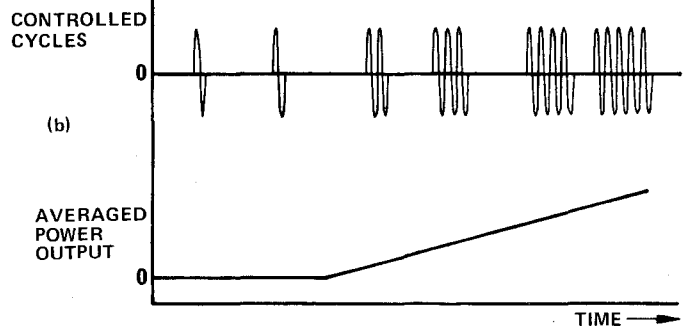
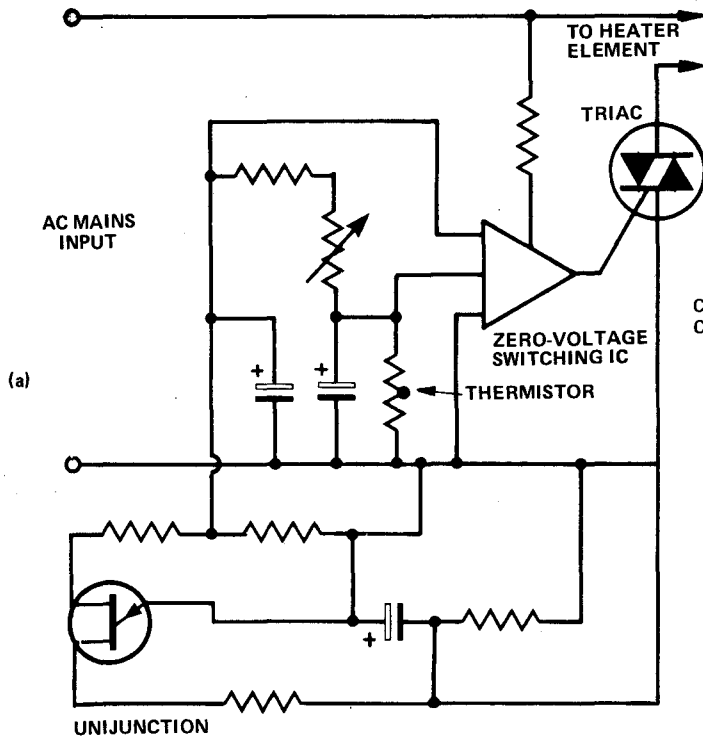


Fig. 7. (a) Zero-voltage-switching temperature controller using a zero voltage switching IC driving a TRIAC from a thermistor sensor. (b) Output signal with gradually increasing power.

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signal to be amplified. This principle is used in high-current dc motor control and in advanced forms of audio amplifier.

INVERTERS AND CONVERTERS

A converter, in the electrical power engineering sense, is a machine (or a circuit) that changes current from one kind to another, or from one frequency to another. An inverter, in the same sense, is a machine that specifically converts dc to ac — being one kind of converter. Originally rotating machines were used but today the trend is to use static solid-state equipment.

There are many instances where these are required — providing a 240 V ac 50 Hz supply when only 12 V batteries exist, providing a 200 V dc supply from 12 V dc and to change frequency such as where a 240 V ac 50 Hz mains might be needed to drive aircraft equipment operating at 400 Hz.

The basic principles used in each are based on the technology discussed before in this part. These are now summarised with examples of the procedures used.

AC to DC: This conversion path has been discussed when we dealt with rectification in Part 14. A transformer is used to obtain the required ac voltage; this is then rectified with diodes and smoothed to provide dc.

DC to AC: This path first changes the dc into a suitable ac signal which can then be transformed to the desired signal level. The frequency of the ac signal is decided by the output load requirement for once produced it must remain at that frequency. (In some cases it is preferable to make use of a higher frequency than 50 Hz).

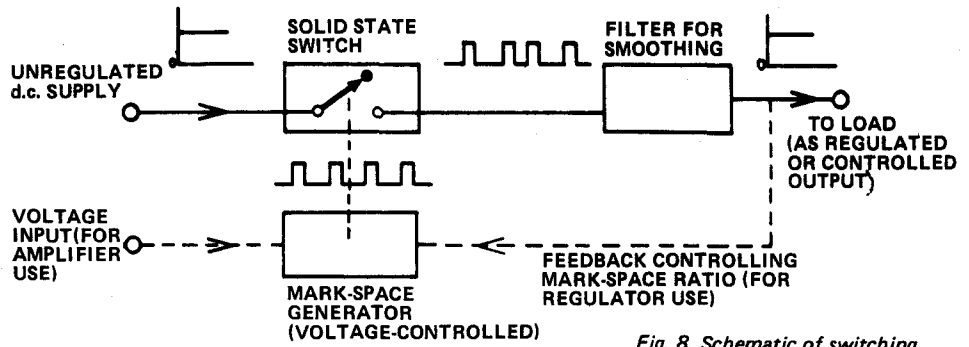


Fig. 8. Schematic of switching regulator or controller (see text).

Figure 10 shows a number of configurations used to produce ac power from a dc supply.

Switching produces square-wave energy after inversion and in many instances this roughly square-wave output waveform is satisfactory. Where the output must be sinusoidal more complex circuitry is required to obtain an undistorted wave shape. When choosing a commercially made inverter it is important to verify if the output waveshape is suitable for the task.

Crystal oscillators can be incorporated into an inverter design where the output frequency must be kept within exacting limits.

DC to DC: The procedure here is to first form the dc to ac conversion. After transformation to the correct voltage (usually the need is a voltage increase) with a double-wound transformer the output is full-wave rectified and smoothed. The transformers used use special iron laminations material to get the best out of the square-wave input waveforms. Figure 11 is a typical 200 W dc to dc up-converter. The transistors, in conjunction with the transformer primary, form a square-

wave oscillator circuit causing flux changes in the transformer which induce the higher output voltage needed for rectification back to dc current.

AC to AC: Some mains equipment can run on either 50 or 60 Hz frequency with little change in performance. Occasionally, however, it is necessary to use the correct frequency specified. To change frequencies the simplest procedure is to convert the original ac supply to a suitable dc value inverting this back to ac at the other frequency. This procedure is easiest to implement because it makes use of standard rectification and inverter packages.

The cost of semiconductor converters has fallen rapidly over the 1970 decade. This has brought about new philosophies in power electrical engineering. In the future there will be more use made of dc electrical transmission. Speed-changing motors are becoming easier to implement using frequency-varied supplies to drive conventional ac machines. Large dc motors are also becoming useful again because regenerative braking of large units — using them as a generator driving into a load — can be put to use

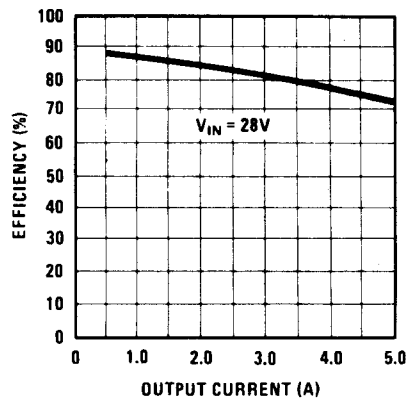
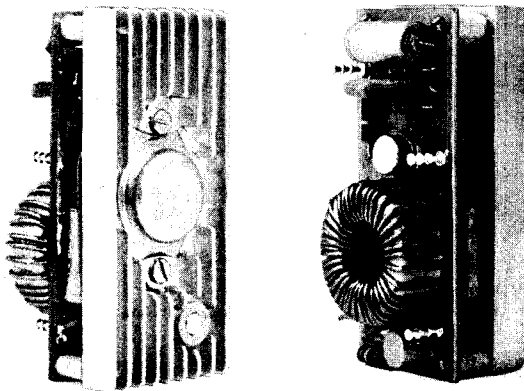


Fig. 9. Switching regulators provide the smallest overall package due to the greatly reduced heatsink needed. They are also the most efficient form of regulator.

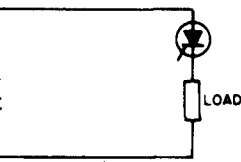
to charge power into the ac mains by the use of dc to ac inverters.

Revolutions have occurred in both power and signal electronics. Attitudes to problem solving are now quite different to just a decade ago. No doubt this trend will continue.

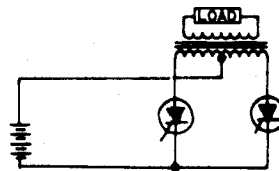
FURTHER READING

Heatsink Design: Several articles were published in *Electronics* October – December, 1973. Most power semiconductor device manufacturers have design sheets available. Heatsink makers usually provide graphs and design procedures. Forced-air cooling is discussed in "Cooling high-power equipment by forced-air convection", L. Katz, *Electronics*, September 21, 1964. A good basic treatment of design is "Heatsinks for power transistors" *Miniwatt Digest*, December, 1962. A deeper discussion is "Design of cooling fins for silicon power rectifiers" *Mullard Tech. Comm. Vol.5*, June, 1960. "Silicon Rectifier Handbook" Motorola, 1966, devotes a chapter to heatsink design.

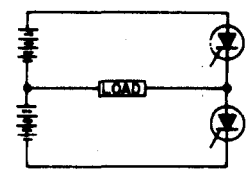
Power Devices The above Motorola handbook is most useful on power diode design. A companion volume



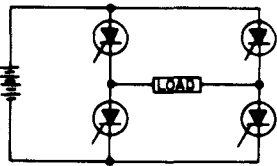
1. CHOPPER



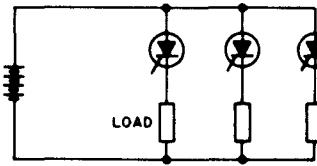
2. CENTER-TAPPED LOAD



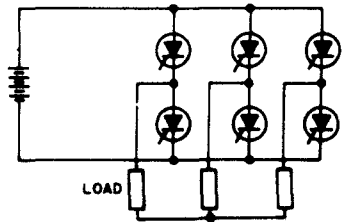
3. CENTER TAPPED SUPPLY



4. BRIDGE



5. THREE PHASE HALF WAVE



6. THREE PHASE BRIDGE

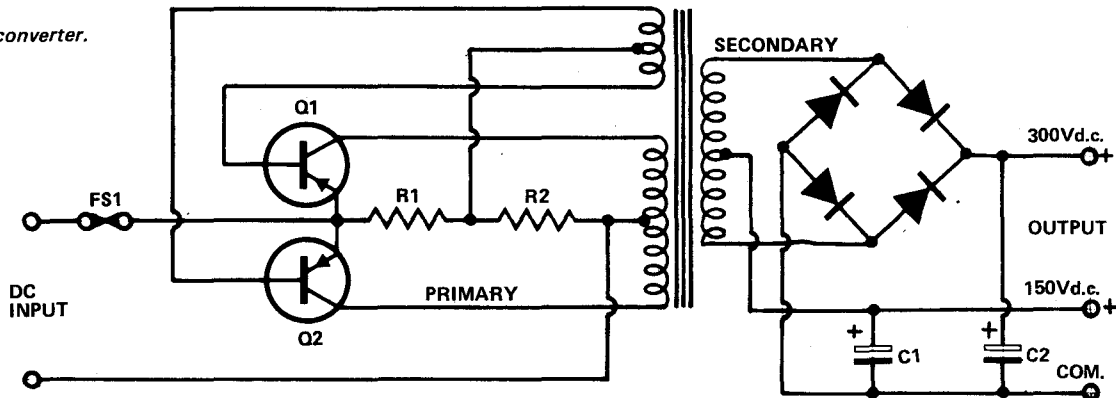
Fig. 10. Various inverter configurations using SCR switches. Triggering methods have been omitted for clarity.

"Power transistor Handbook 1961" deals with transistors. SCRs are covered in "SCR Manual", General Electric, 1964.

Converters and Inverters: An early paper "Self-excited inverters" W.J.R. Farmer and R.J. Spreadbury, *AFI Engng.*, March, 1965 provides a tutorial summary. Many power engineering companies can provide papers and articles.

Switching regulators: Low to medium power units based on an IC are discussed in "Designing switching regulators" R.J. Widlar, *National Semiconductor AN-2*, 1969. See also "Switching regulators: the efficient way to power", R.S. Olla, *Electronics*, August 16, 1973, and application notes of Motorola, Delco Radio, RCA, Kepco, and others.

Fig. 11 Circuit of dc up-converter.



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Volume 3, which also contains a comprehensive index for the complete series, will be published in Nov/Dec this year.

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The famous Sound of Koss takes on a new look.



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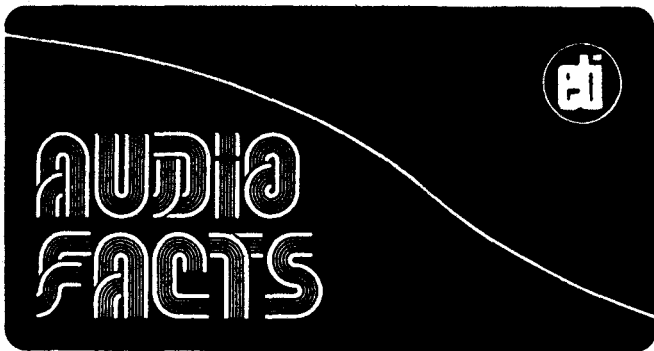
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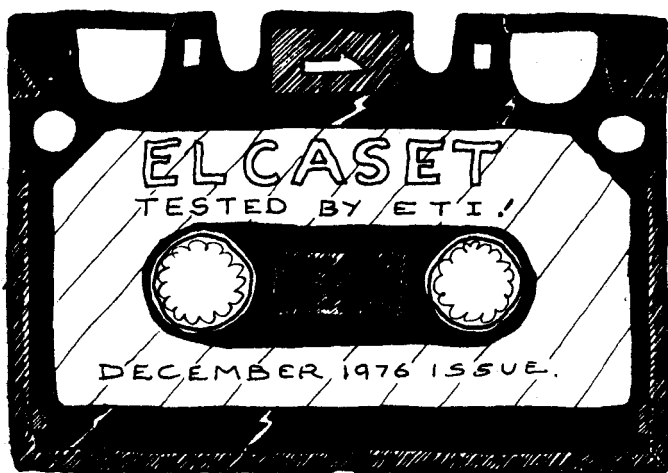
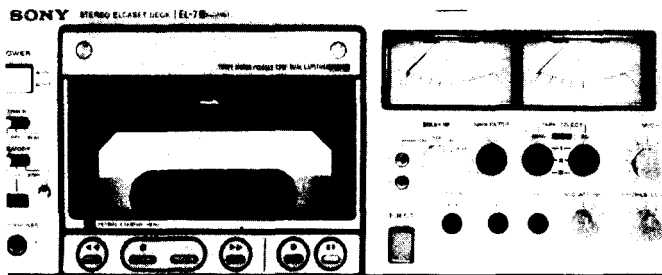


ELCASET

The new large format Eicaset system described in Audio Facts last month really is for real. We know because we've had one for appraisal throughout the past month.

How well does it work?

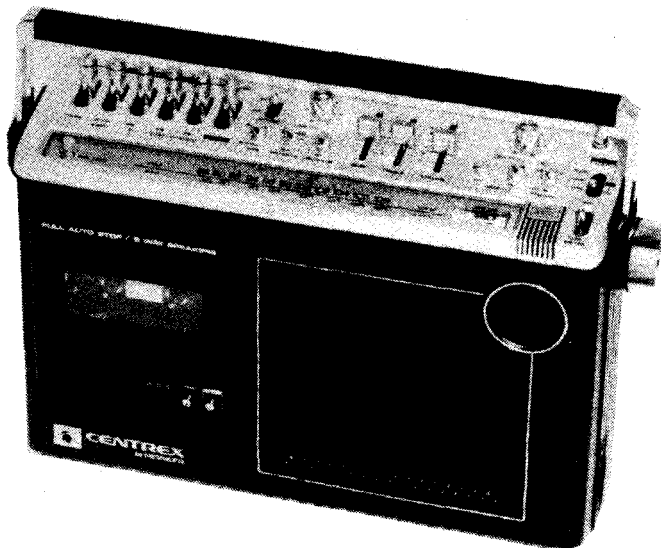
Next month's ETI will carry a full review and technical description of this fascinating new audio development.



AUSTRALIAN CES - 1977

Following the outstanding success of the 1st Australian Consumer Electronics Show held recently at the Sydney Hilton, the organisers, Riddell Exhibitions advise us that plans are now well ahead for a similar but larger show in 1977.

Riddells advise us that the 1977 show will again be held at the Sydney Hilton from Thursday 4th August through Tuesday 9th August. The Thursday, Friday and Saturday will be public days - Sunday, Monday and Tuesday will be for the trade.



Pioneer recently announced the launch of a new division - Centrex - and with it a new range of audio products under the name 'Centrex by Pioneer'.

The initial Centrex product release consists of three radio cassette recorders and a complete home stereo system.

IREE AUDIO GROUP

This month is the fourth anniversary of the Audio Group sponsored by IREE Australia.

Since its beginning in November 1972, the group has since grown to well over 100 active participating members. The Group was formed to promote and serve the special interests of both IREE members and others in the areas of audio-frequency engineering and technology. Membership is open to all person interested in audio.

Regular meetings include lectures and demonstrations on all audio topics and frequent technical tours of industrial and scientific establishments involved in the audio field.

Details of membership from

P. Garde, Hon Secretary, School of Electrical Engineering University of NSW P.O. Box 1 Kensington, NSW 2033

LOW DISTORTION OSCILLATOR FOR CHECKING HI-FI EQUIPMENT

The problem of measuring distortion in equipment which itself has very low distortion places considerable importance on the distortion present in signal sources themselves. The Meguro Denpa MCR-402 Low Distortion Oscillator has been developed for use as a test source in such applications.

Output distortion in the range from 100 Hz to 10 kHz is less than 0.01% and only rises to 0.1% at 100 kHz. Three decade switches and a multiplier set frequencies from 0.2 Hz to 100 kHz. A fine control for closer adjustment of output is also provided.

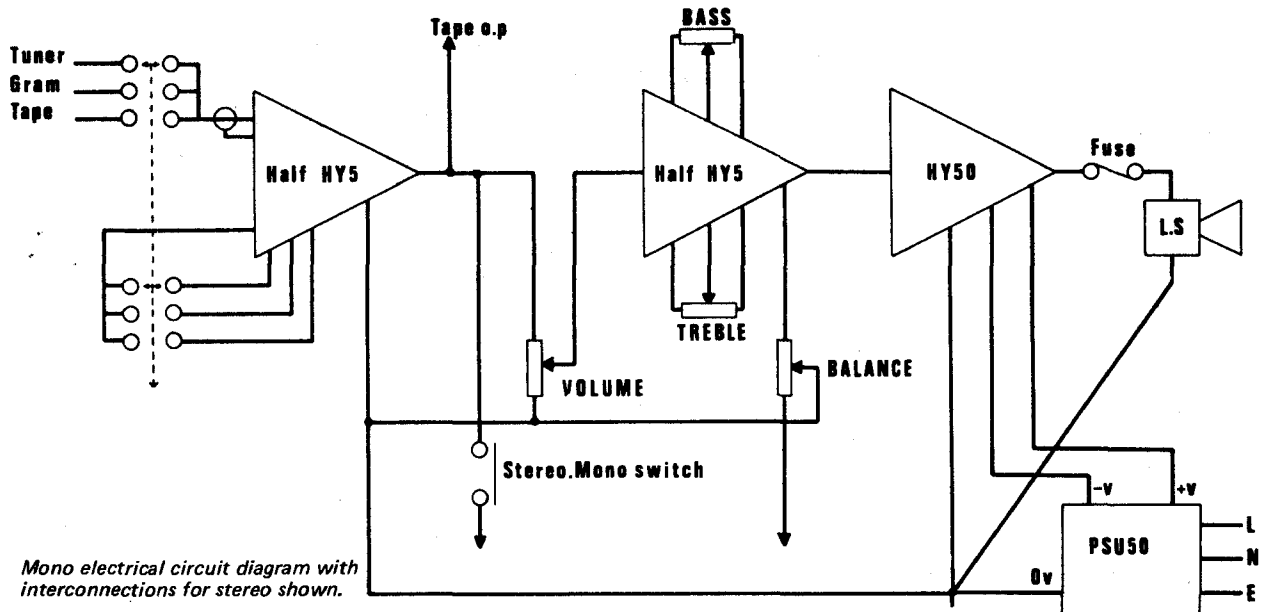
A calibrated meter provides output level indication and the level can be adjusted over a wide range by a stepped attenuator with fine control. In addition to the main output a quadrature output is also provided.

Frequency can be synchronised with an external source where higher accuracy or stability are called for. It is also possible to use the MCR402 as a "convertor" for distorted signals.

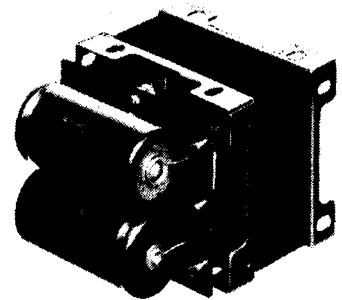
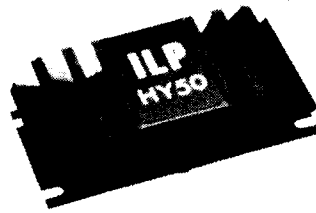
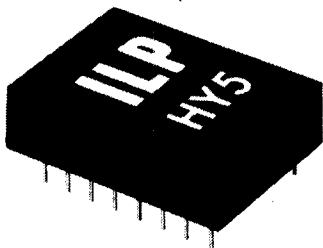
Full information on the Meguro Denpa MCR-402 and other specialised broadcast test equipment is available from Parameters Pty Ltd. 68 Alexander Street, Crows Nest 2065.



SHEER SIMPLICITY!



Mono electrical circuit diagram with interconnections for stereo shown.



The HY5 is a complete mono hybrid preamplifier, ideally suited for both mono and stereo applications. Internally the device consists of two high quality amplifiers—the first contains frequency equalisation and gain correction, while the second caters for tone control and balance.

TECHNICAL SPECIFICATION

Inputs
 Magnetic Pick-up 3mV, RIAA
 Ceramic Pick-up 30mV
 Microphone 10mV
 Tuner 100mV
 Auxiliary 3-100mV
 Input impedance 47kΩ at 1kHz.

Outputs
 Tape 100mV
 Main output 0db (0.775 volts RMS)

Active Tone Controls
 Treble ±12db at 10kHz
 Bass ±12db at 100Hz

Distortion 0.05% at 1kHz
Signal/Noise Ratio 68db
Overload Capability 40db on most sensitive input

Supply Voltage ±16-25 volts.
 PRICE \$16.06 P&P \$0.30

The HY50 is a complete solid state hybrid Hi-Fi amplifier incorporating its own high conductivity heatsink hermetically sealed in black epoxy resin. Only five connections are provided: Input, output, power lines and earth.

TECHNICAL SPECIFICATION

Output Power 25 watts RMS into 8Ω
Load Impedance 4-16Ω
Input Sensitivity 0db (0.775 volts RMS)
Input Impedance 47kΩ
Distortion Less than 0.1% at 25 watts typically 0.05%
Signal/Noise Ratio Better than 75db
Frequency Response 10Hz-50kHz ±3db
Supply Voltage ±25 volts
Size 105 x 50 x 25 mm.

PRICE \$20.27 P&P \$0.40

P&P \$2.00 FOR 1 COMPLETE SET OF HY5 + HY50 + PSU50

The PSU50 incorporated a specially designed transformer and can be used for either mono or stereo systems.

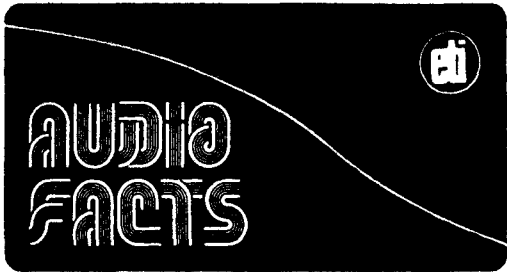
TECHNICAL SPECIFICATIONS

Output voltage 50 volts (25-0-25)
Input voltage 210-240 volts
Size L.70, D.90, H.60 mm.
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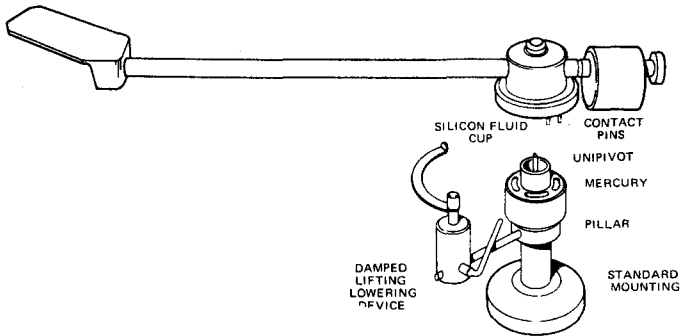
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 Enclose Cheque Postal Orders Money Order
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KMAL LABORATORY PICK-UP ARM

In past few years a number of companies have produced pick-up cartridges having very low mass and high compliance and designed to track at pressures of a gram or so.

In many cases purchasers of these cartridges have run into problems due to the large mass and correspondingly high inertia of the conventional pick-up arms to which the cartridges were fitted.



Basically a pick-up arm must have low effective mass — that is the moving bit as seen by the stylus must be light, the fundamental resonance of the arm must not coincide with audio frequencies, nor frequencies of record surface irregularities nor motor rumble; the arm must have low friction and inertia so that little load is applied to the record groove, and so that the stylus may faithfully follow warped records; and the arm must be adequately damped and accurate anti-skating bias provided.

These requirements are hard to meet. Most high quality arms are in fact very complex engineering structures indeed.

The KMAL arm produced by Britain's Keith Monks Audio solves most of the problems inherent in arm design in a simple and elegant fashion. The arm is a unipivot design intended specifically for use with low mass ultra-high compliance cartridges. The arm itself is pivoted on an inverted needle, the point of which sits in the centre of a precision ball race. A metal collar protruding downwards from the race rests in a small container of a viscous silicone fluid.

Unlike other arms, the audio output from the cartridge is transferred from the arm assembly via four metal pins immersed in tiny baths of mercury. Anti-skating bias is applied via a simple magnetic device acting on the pins.

Brief tests showed that the stiction and friction in both planes was virtually identical at 3-4 mg — about the lowest we've ever seen quoted before.

Frankly the KMAL arm won't win any prizes for appearance — it's not badly made but it's not a patch on the SME (for example), to look at. Aesthetics aside however it does work very well indeed and will no doubt appeal to those who have an affinity for its totally straightforward and elegant engineering approach to solving fundamental problems.

Further details may be obtained from W.C. Wedderspoon Pty Ltd., 3 Ford St Greenacre NSW 2190

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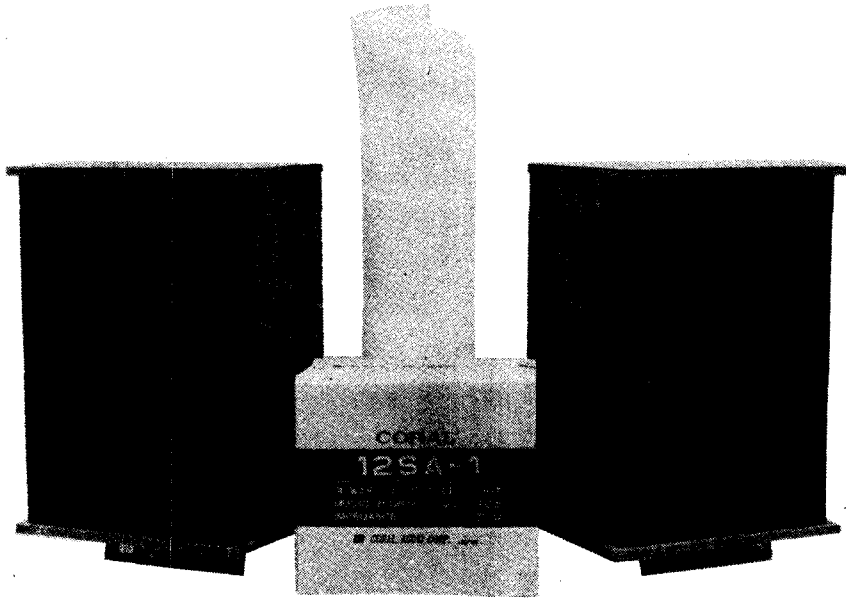
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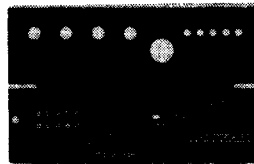
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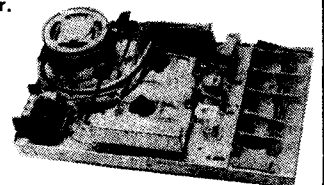
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Ideas for experimenters

These pages are intended primarily as a source of ideas. As far as reasonably possible all material has been checked for feasibility, component availability etc, but the circuits have not necessarily been built and tested in our laboratory. Because of the nature of the information in this section we cannot enter into any correspondence about any of the circuits, nor can we produce constructional details. Electronics Today is always seeking material for these pages. All published material is paid for — generally at a rate of \$5 to \$7 per item.

Extra facilities on the HP21

Although the Hewlett-Packard HP21 is very versatile, it does lack a key which allows the contents of the memory to be interchanged with those of the x register. This can be achieved in the following manner:—

First press the blue function key once. Then press the 'minus' key and the '7' key, at the same time,

*i.e., Step 1 (F)
Step 2 (-) (7) simultaneously*

This should interchange the number stored in the memory with that in the display.

Also if pressing the function key and

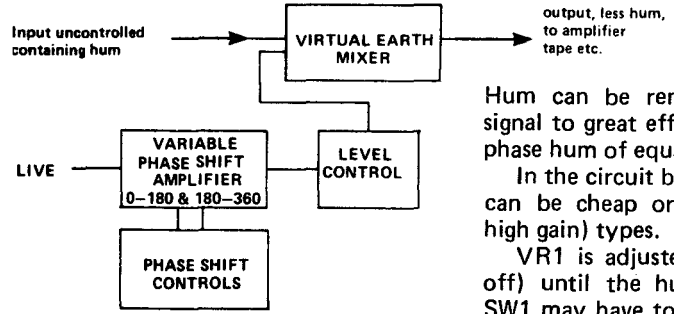
then the π key seems a bit slow to obtain π , pressing (\div) and (0) simultaneously will yield the constant. With a bit of practice this can be done quicker than the conventional way.

Finally a number can have its inverse calculated and stored in the memory while the number which was previously in the memory is now displayed, merely by pressing the function key and then simultaneously pressing (1/x) and (cos). However, this feature is probably not very important.

So far I know of no way to get a timer as can be obtained on the HP45.

Hum stopper

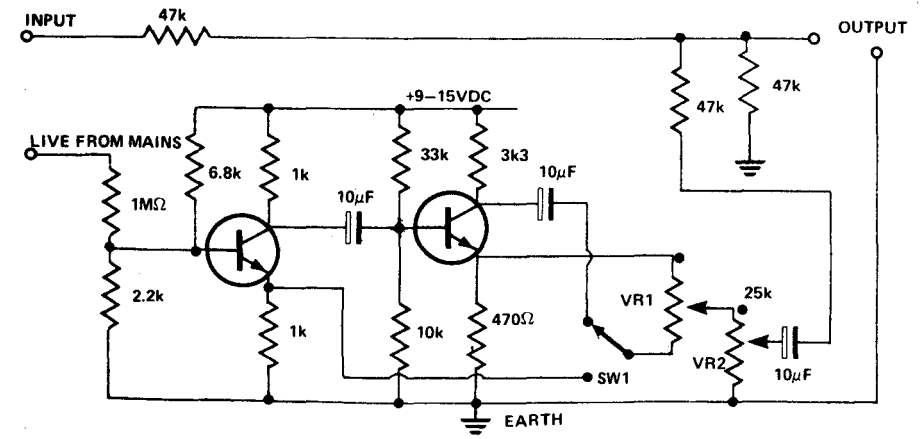
BLOCK DIAGRAM OF SYSTEM



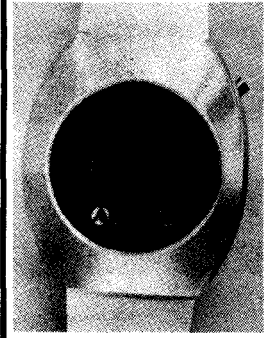
Hum can be removed from an audio signal to great effect by mixing an anti-phase hum of equal level.

In the circuit below all the transistors can be cheap or surplus npn (low or high gain) types.

VR1 is adjusted with VR2 low (not off) until the hum is at a minimum, SW1 may have to be changed over, then the level VR2, is altered until the hum is removed.



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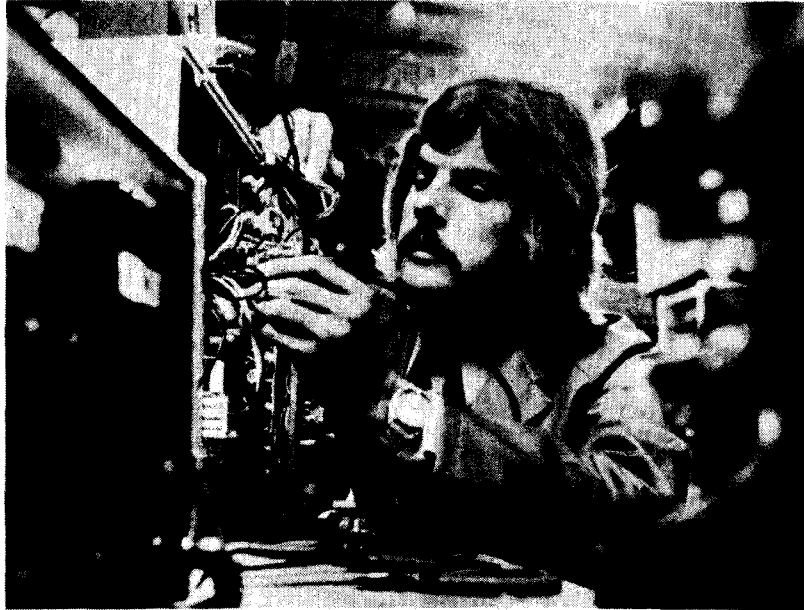
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COLOUR TV

If you already have a knowledge of the principles and practice of TV, this course will prepare you for the introduction of colour TV. Subjects covered in detail include: Colour in TV, the Colour TV system, Picture Tubes and Receiver Circuits for Colour TV, Troubleshooting Colour TV, Alignment of Monochrome and Colour Receivers and the PAL System.

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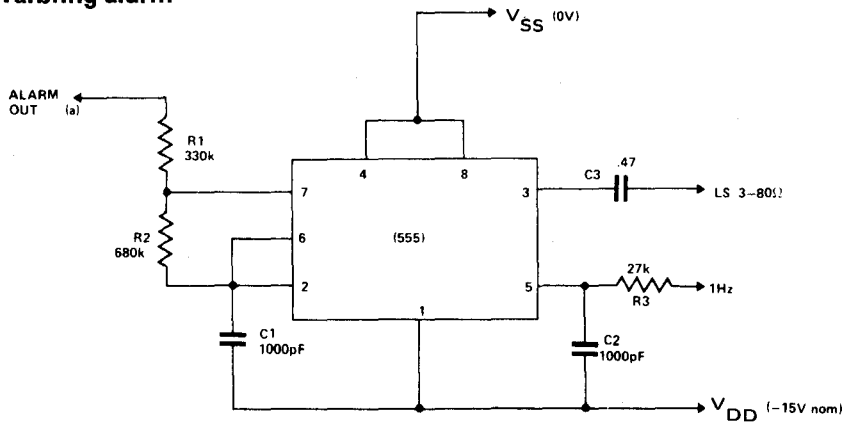
EAO

POST THIS COUPON TODAY TO

ICS

Ideas for experimenters

Warbling alarm

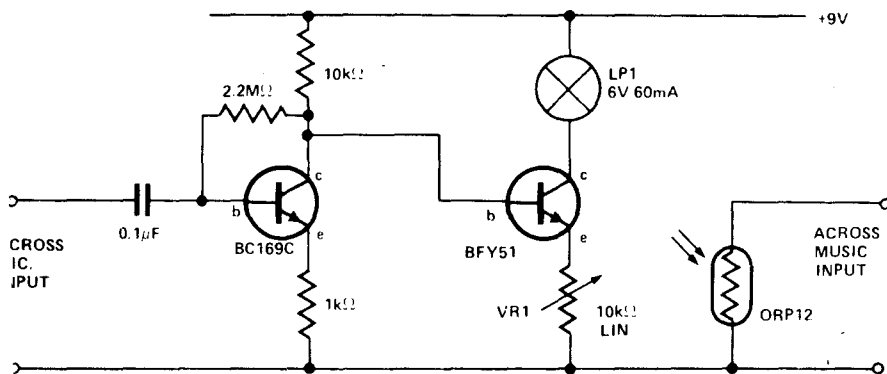


This device gives a two-tone alarm from a digital clock. It may be used with any CMOS alarm clock chip having an active high alarm output and 1 Hz (optional) output. It was built to work with a CT7001 chip and requires no interface components.

The 555 operates in normal astable mode when the alarm goes high (ie point [a] approaches V_{SS}). Pin 5 is the normal control voltage input and swings from almost V_{SS} to V_{DD} via the 27 k resistor at a 1 Hz rate. This causes the

audio output to switch between high and low tones, above and below the frequency determined by R1, R2 and C1. To vary the frequency difference, R3 may be altered within wide limits, but it is inadvisable to keep it below 15 k. The basic frequency is best varied by changing C1. Audio output may be varied by changing C3 (depending on LS impedance). In the original, a 35 Ω speaker was used with -12 V V_{DD} and was sufficient to rouse an expert heavy sleeper.

Simple disco auto fade



When a DJ has to make an announcement over a record, the normal procedure is to fade out the deck, fade in the microphone, and vice versa at the end. If this unit is used, however, the operator need only speak into the microphone, the deck being faded out automatically.

The lamp and LDR need to be taped together, preferably with black tape, to exclude light. VR1 is used to set the

brightness of the lamp. With no signal on the input, VR1 is set for no attenuation of the music signal. When speaking at normal volume through the mic, the music should fade down until it can be heard quietly in the background. Some microphones may not produce enough signal to do this. If this happens, a simple pre-amp can be added to the input stage, as was done in the prototype.

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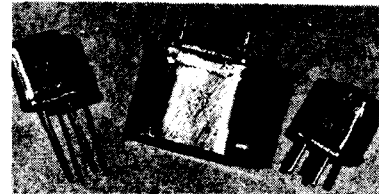
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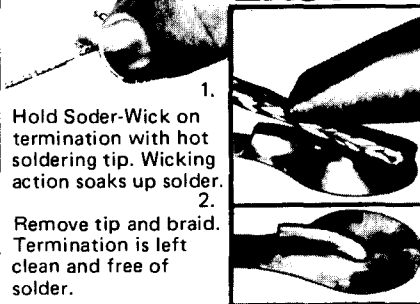
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We get many enquiries from readers wanting to know where they can get kits for the projects we publish. The list below indicates the suppliers we know about and the kits they do. These kits include hardware as well as electronic components. There are many suppliers who can provide you with all the electronics for a project (companies like Techniparts in Brisbane) so it is not necessary to buy a kit provided you can find a suitable case etc.

Any companies who want to be included in this list should phone Steve Braidwood on 33-4282.

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ETI 102 Audio Signal Generator E,D
ETI 103 Logic Probe. E
ETI 107 Widerange Voltmeter E
ETI 108 Decade Resistance Box E
ETI 109 Digital Frequency Meter E
ETI 111 IC Power Supply E
ETI 112 Audio Attenuator E
ETI 113 7-Input Thermocouple Meter. E
ETI 116 Impedance Meter E
ETI 117 Digital Voltmeter E,A
ETI 118 Simple Frequency Counter E,A
ETI 119 5V switching regulator supply E
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ETI 121 Logic Pulsar E
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ETI 422B Booster Amp E

Key to the companies:

- A Applied Technology Pty. Ltd. of Hornsby, NSW.
- C Amateur Communications Advancements, PO Box 57, Rozelle, NSW.
- D Dick Smith Pty. Ltd. of Crows Nest, NSW.
- E E.D. & E. Sales, Victoria.
- J Jaycar Pty. Ltd. of Haymarket, NSW.
- N Nebula Electronics Pty. Ltd. of Rushcutters Bay, NSW.

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two-decade, auto-ranging, digital meter. **Readout Accuracy:** Voltage: Analog $\pm 3\%$ of rated output; Digital $\pm 0.5\%$ of reading ± 1 count using lab standards, $\pm 1.0\%$ of reading ± 1 count using built-in calibrator, Current: Analog $\pm 3\%$ of rated output; Digital $\pm 1.0\%$ of reading ± 4 count using lab standards, $\pm 1.5\%$ of reading ± 4 count using built-in calibrator. **Readout Response Time: (Digital):** 2 seconds to within 5 counts. **Stability at Output Terminals:** Voltage: $\pm(0.01\% + 1 \text{ mV})/\text{hr}$. Current: $\pm(0.05\% + 1 \text{ mA})/\text{hr}$. **Stability as Displayed (Digital):** Voltage: $\pm(0.01\% + 1 \text{ mV} + \frac{1}{2} \text{ count})/\text{hr}$. Current: $\pm(0.01\% + 1 \text{ mA} + \frac{1}{2} \text{ count})/\text{hr}$. **Operating Modes:** Constant voltage, constant current, auto-series, auto parallel. **Programming Mode:** Voltage: A—zero to rated output with 0 to 5.0 VDC applied; B—zero to rated output with 0 to 5000Ω external resistor. Current: Zero to rated output with applied voltage of 1.0 volt/amp (0.1 volt/amp for IP-2731). Programming transient response: 0.1 ms for low current to high current changes: 1.0 ms for high to low. Load resistance less than $10 \times$ (E Rated/I Rated). **Power Requirements:** 120/240 $\pm 10/20$ VAC; 60/50 Hz, 2.0/1.0 Amps max. **Overall Dimensions:** 5.5" H x 15" W x 13.5" D.

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422 Power Amp

In order to adjust the bias current on a 422 amplifier I connected 220 ohm resistors across the fuseholders and switched on the power. I get a full 40 V across the resistor in the positive side and no voltage across the resistor on the negative side. The presets RV1 and RV2 have no effect. Can you help me find the fault, please?

G.M., Manly, NSW.

These symptoms indicate that one of the output transistors is fully turned on and the other is fully off. This suggests you have inserted one of the components incorrectly. Check the BC548s and make sure that they are Philips types — if they are from another manufacturer they must be inserted in a different way to that shown on the overlay.

4k7 = 47 k or 4.7 k?

In the parts list of ETI project 122 R7 is quoted as 4k7. It is also coded this way on the circuit diagram. Shouldn't this be 47 k?

H.G.M., Danganong, Victoria.

When these codings are used it is in conformity with an international standard devised to overcome certain printing errors (the obliteration or erroneous inclusion of the decimal point). In these examples the multiplier occupies the position normally taken by the decimal point. So 4k7 is 4.7 k, 3M9 is 3.9 M, 5n6 is 5.6 nF or 0.0056 μ F. We have decided to adopt this standard for all future ETI projects.

Toroidal transformers

Recent hi-fi advertisers in ETI have mentioned that their amplifiers have toroidal power transformers. What does this mean, and what advantages do these types offer?

D.J., Melbourne.

Toroidal coils and transformers are wound on a ring-shaped core made from compressed alloy dust. Their advantages lie in the high Q values obtainable (as much as several hundreds). Toroidal coils are used in equalisers because of the high efficiency — an added advantage is that they are largely unaffected by external magnetic fields so they can be mounted close together.

When used in hi-fi amplifiers as mains transformers they have two advantages. First, the closed magnetic circuit keeps down hum fields, and second, the flat construction means that a low-profile case design can be used for the amplifier.



This new feature is our response to the many requests we get from readers who want explanation or information on topics they read about in the magazine. If you have a question please send it to Please Explain, ETI Magazine, 15 Boundary Street, Rushcutters Bay, NSW. 2011.

Remote Control Receiver

Two questions about this project —

- (1) Does it matter if the receiver crystal is 455 kHz lower than the transmitter frequency (instead of 455 kHz higher)?
- (2) Why does IC3 (4001 CMOS) have to be an SCL4001 or an HEF4001? Why not any old 4001 CMOS IC?

H.A., Mt Waverley, Vic.

(1) It doesn't matter.

(2) The Signetics and Solid-State Scientific devices must be used because they are faster than the usual CMOS.

K or k?

In ETI I have seen references to memory chips with 1K bytes capacity and on other occasions I've seen the expression 1k used. Is there any difference?

K.D., Sydney.

A memory with 1K byte storage capacity has room to store 1024 words. This number is two to the power of ten and it is symbolised by the capital letter K (when used to specify large numbers in computer articles). Rather than referring to numbers like 65 536 (2¹⁶) we say 64K (or 65k).

When the small k is used as a multiplier it introduces a factor of 1000, as in kg, km, kW, etc. Sometimes it is used *incorrectly* to indicate 2¹⁰.

Expander Compressor

In project ETI 443 I can't work out how the connections between the IC sockets and the pcb tracks on the component side of the board are made.

Also could you put me straight as to where the unit connects into the audio chain?

R.D.D., Bendigo, Victoria.

Ordinary IC sockets cannot be used with the double-sided pcb design published for this project. If you really want to use sockets they need to have extra-long pins (try using wire-wrap types).

The Expander-Compressor needs to be connected at a level of at least 100 mV to work properly. It should go between the preamp and the main amplifier and ideally it would come after the volume control. Connection between the Tape Out and Tape In sockets would probably be satisfactory.

Digital Display DPs

On the ETI 118 frequency counter project the range-switch wiring diagram refers to DP1 and DP2. What are these connections?

A.S., Sydney.

These refer to the decimal points on the seven-segment displays (actually on display 1 and display 2) of the digital display module.

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PENDING MOVE necessitates selling majority of Hi-fi gear - Junk box parts. Hundreds of items for Hobyist. G. Sankowsky (Butterfly farm), Long Rd., Mt. Tamborine, Q. Ph. 075-45-1353.

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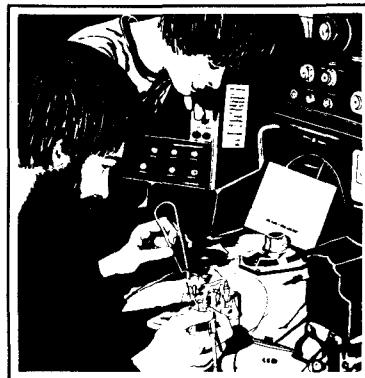


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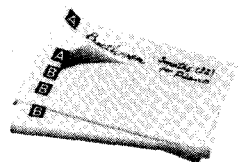


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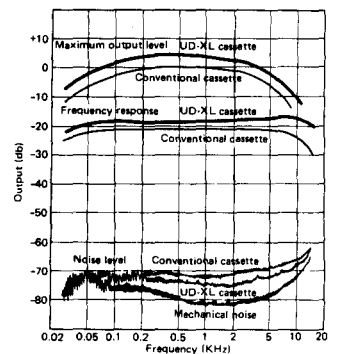
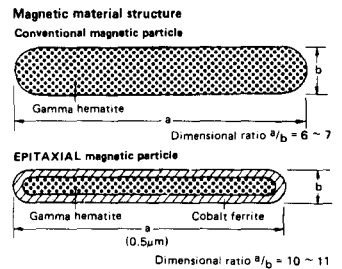
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Another good idea of the UD-XL cassette is a replaceable self-index label. Simply peel off the old label and put on a new one when you change the recording contents. No more mess on the label.

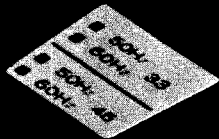
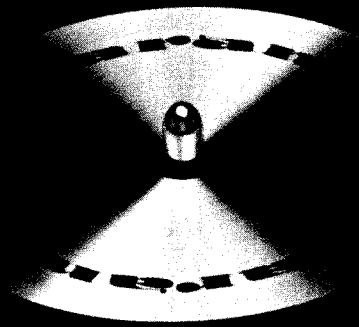


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For further information please write to Maxell Advisory Service, P.O. Box 49, Kensington, N.S.W. 2033.

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